

Freescale Semiconductor

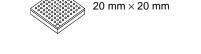
Data Sheet:

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MSC8122

Quad Digital Signal Processor

- Four StarCore[™] SC140 DSP extended cores, each with an SC140 DSP core, 224 Kbyte of internal SRAM M1 memory (1436 Kbyte total), 16 way 16 Kbyte instruction cache (ICache), four-entry write buffer, external cache support, programmable interrupt controller (PIC), local interrupt controller (LIC), and low-power Wait and Stop processing modes.
- 475 Kbyte M2 memory for critical data and temporary data buffering.
- 4 Kbyte boot ROM.
- M2-accessible multi-core MQBus connecting the M2 memory with all four cores, operating at the core frequency, with data bus access of up to 128-bit reads and up to 64-bit writes, central efficient round-robin arbiter for core access to the bus, and atomic operation control of M2 memory access by the cores and the local bus.
- Internal PLL configured are reset by configuration signal values.
- 60x-compatible system bus with 64 or 32 bit data and 32-bit address bus, support for multi-master designs, four-beat burst transfers (eight-beat in 32-bit data mode), port size of 64/32/16/8 bits controlled by the internal memory controller, access to external memory or peripherals, access by an external host to internal resources, slave support with direct access to internal resources including M1 and M2 memories, and on-device arbitration for up to four master devices.
- Direct slave interface (DSI) using a 32/64-bit slave host interface with 21–25 bit addressing and 32/64-bit data transfers, direct access by an external host to internal and external resources, synchronous or asynchronous accesses with burst capability in synchronous mode, dual or single strobe mode, write and read buffers to improve host bandwidth, byte enable signals for 1/2/4/8-byte write granularity, sliding window mode for access using a reduced number of address pins, chip ID decoding to allow one \overline{CS} signal to control multiple DSPs, broadcast mode to write to multiple DSPs, and big-endian/little-endian/munged support.
- Three mode signal multiplexing: 64-bit DSI and 32-bit system bus, 32-bit DSI and 64-bit system bus, or 32-bit DSI and 32-bit system bus.
- Flexible memory controller with three UPMs, a GPCM, a page-mode SDRAM machine, glueless interface to a variety of memories and devices, byte enables for 64- or 32-bit bus widths,



FC PBGA-431

8 memory banks for external memories, and 2 memory banks for IPBus peripherals and internal memories.

- Multi-channel DMA controller with 16 time-multiplexed single channels, up to four external peripherals, DONE or DRACK protocol for two external peripherals, service for up to 16 internal requests from up to 8 internal FIFOs per channel, FIFO generated watermarks and hungry requests, priority-based time-multiplexing between channels using 16 internal priority levels or round-robin time-multiplexing between channels, flexible channel configuration with connection to local bus or system bus, and flyby transfer support that bypasses the FIFO.
- Up to four independent TDM modules with programmable word size (2, 4, 8, or 16-bit), hardware-base A-law/µ-law conversion, up to 128 Mbps data rate for all channels, with glueless interface to E1 or T1 framers, and can interface with H-MVIP/H.110 devices, TSI, and codecs such as AC-97.
- Ethernet controller with support for 10/100 Mbps MII/RMII/SMII including full- and half-duplex operation, full-duplex flow controls, out-of-sequence transmit queues, programmable maximum frame length including jumbo frames and VLAN tags and priority, retransmission after collision, CRC generation and verification of inbound/outbound packets, address recognition (including exact match, broadcast address, individual hash check, group hash check, and promiscuous mode), pattern matching, insertion with expansion or replacement for transmit frames, VLAN tag insertion, RMON statistics, local bus master DMA for descriptor fetching and buffer access, and optional multiplexing with GPIO (MII/RMII/SMII) or DSI/system bus signals lines (MII/RMII).
- UART with full-duplex operation up to 6.25 Mbps.
- Up to 32 general-purpose input/output (GPIO) ports.
- I²C interface that allows booting from EEPROM devices.
- Two timer modules, each with sixteen configurable 16-bit timers.
- Eight programmable hardware semaphores.
- Global interrupt controller (GIC) with interrupt consolidation and routing to INT_OUT, NMI_OUT, and the cores; thirty-two virtual maskable interrupts (8 per core) and four virtual NMI (one per core) that can be generated by a simple write access.
- Optional booting external memory, external host, UART, TDM, or I²C.





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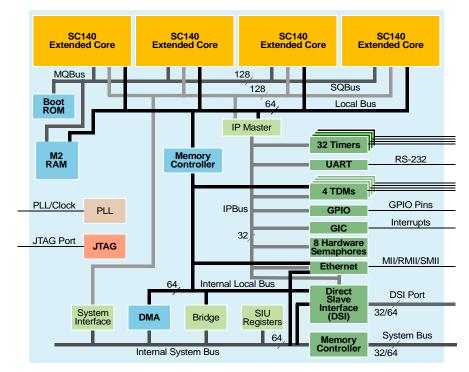
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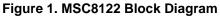
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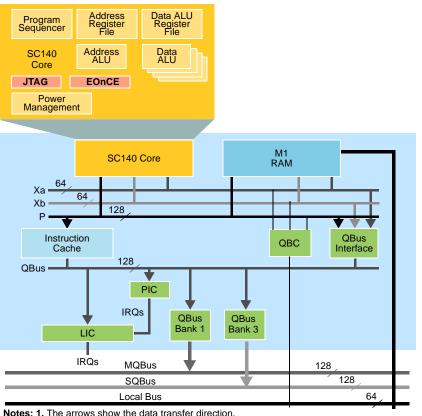
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Package









Notes: 1. The arrows show the data transfer direction.
2. The QBus interface includes a bus switch, write buffer, fetch unit, and a control unit that defines four QBus banks. In addition, the QBC handles internal memory contentions.

Figure 2. StarCore SC140 DSP Extended Core Block Diagram



ssignments

1 Pin Assignments

This section includes diagrams of the MSC8122 package ball grid array layouts and pinout allocation tables.

1.1 FC-PBGA Ball Layout Diagrams

Top and bottom views of the FC-PBGA package are shown in Figure 3 and Figure 4 with their ball location index numbers.

MSC8122 Quad Digital Signal Processor Data Sheet, Rev. 16

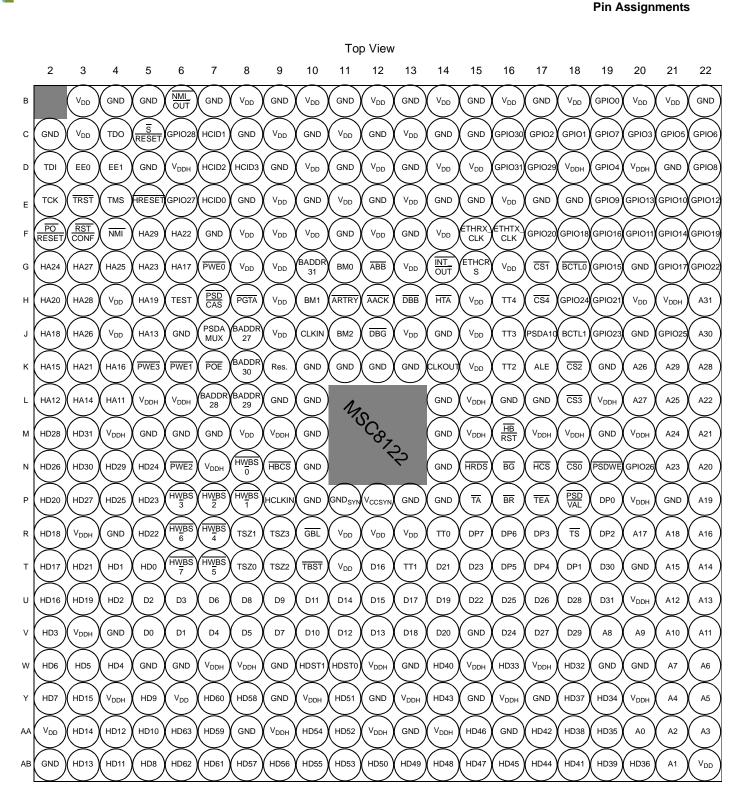


Figure 3. MSC8122 Package, Top View



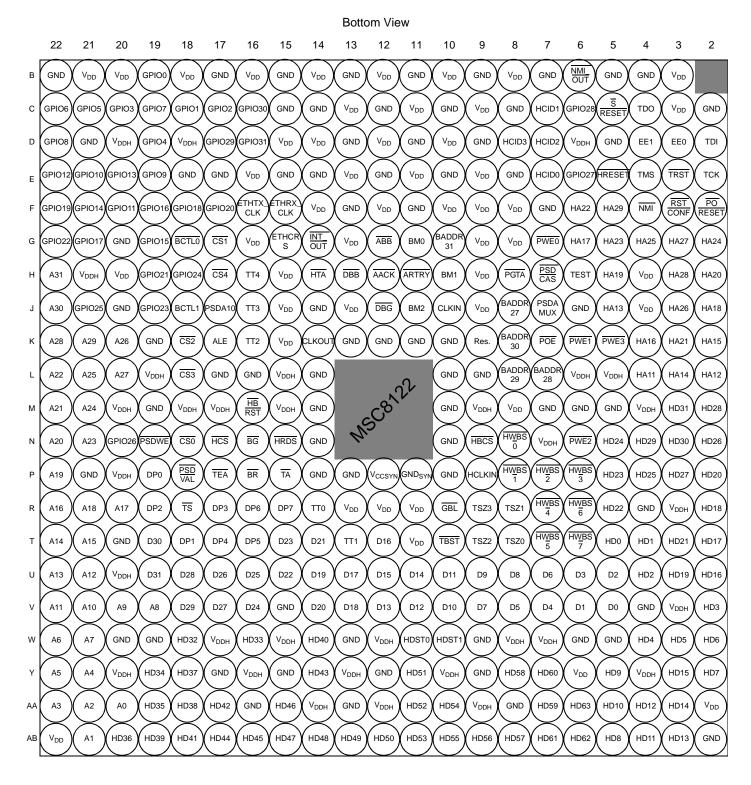


Figure 4. MSC8122 Package, Bottom View

1.2 Signal List By Ball Location

 Table 1 presents signal list sorted by ball number.

```
Table 1. MSC8122 Signal Listing by Ball Designator
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Des.	Signal Name	Des.	Signal Name		
B3	V _{DD}	C18	GPIO1/TIMER0/CHIP_ID1/IRQ5/ETHTXD1		
B4	GND	C19	GPIO7/TDM3RCLK/IRQ5/ETHTXD3		
B5	GND	C20	GPIO3/TDM3TSYN/IRQ1/ETHTXD2		
B6	NMI_OUT	C21	GPIO5/TDM3TDAT/IRQ3/ETHRXD3		
B7	GND	C22	GPIO6/TDM3RSYN/IRQ4/ETHRXD2		
B8	V _{DD}	D2	TDI		
B9	GND	D3	EE0		
B10	V _{DD}	D4	EE1		
B11	GND	D5	GND		
B12	V _{DD}	D6	V _{DDH}		
B13	GND	D7	HCID2		
B14	V _{DD}	D8	HCID3/HA8		
B15	GND	D9	GND		
B16	V _{DD}	D10	V _{DD}		
B17	GND	D11	GND		
B18	V _{DD}	D12	V _{DD}		
B19	GPIO0/CHIP_ID0/IRQ4/ETHTXD0	D13	GND		
B20	V _{DD}	D14	V _{DD}		
B21	V _{DD}	D15	V _{DD}		
B22	GND	D16	GPIO31/TIMER3/SCL		
C2	GND	D17	GPIO29/CHIP_ID3/ETHTX_EN		
C3	V _{DD}	D18	V _{DDH}		
C4	TDO	D19	GPIO4/TDM3TCLK/IRQ2/ETHTX_ER		
C5	SRESET	D20	V _{DDH}		
C6	GPIO28/UTXD/DREQ2	D21	GND		
C7	HCID1	D22	GPIO8/TDM3RDAT/IRQ6/ETHCOL		
C8	GND	E2	ТСК		
C9	V _{DD}	E3	TRST		
C10	GND	E4	TMS		
C11	V _{DD}	E5	HRESET		
C12	GND	E6	GPIO27/URXD/DREQ1		
C13	V _{DD}	E7	HCID0		
C14	GND	E8	GND		
C15	GND	E9	V _{DD}		
C16	GPIO30/TIMER2/TMCLK/SDA	E10	GND		
C17	GPIO2/TIMER1/CHIP_ID2/IRQ6	E11	V _{DD}		



ssignments

Des.	Signal Name	Des.	Signal Name
E12	GND	G6	HA17
E13	V _{DD}	G7	PWE0/PSDDQM0/PBS0
E14	GND	G8	V _{DD}
E15	GND	G9	V _{DD}
E16	V _{DD}	G10	IRQ3/BADDR31
E17	GND	G11	BM0/TC0/BNKSEL0
E18	GND	G12	ABB/IRQ4
E19	GPIO9/TDM2TSYN/IRQ7/ETHMDIO	G13	V _{DD}
E20	GPIO13/TDM2RCLK/IRQ11/ETHMDC	G14	IRQ7/INT_OUT
E21	GPIO10/TDM2TCLK/IRQ8/ETHRX_DV/ETHCRS_DV/NC	G15	ETHCRS/ETHRXD
E22	GPIO12/TDM2RSYN/IRQ10/ETHRXD1/ETHSYNC	G16	V _{DD}
F2	PORESET	G17	CS1
F3	RSTCONF	G18	BCTL0
F4	NMI	G19	GPIO15/TDM1TSYN/DREQ1
F5	HA29	G20	GND
F6	HA22	G21	GPIO17/TDM1TDAT/DACK1
F7	GND	G22	GPIO22/TDM0TCLK/DONE2/DRACK2
F8	V _{DD}	H2	HA20
F9	V _{DD}	H3	HA28
F10	V _{DD}	H4	V _{DD}
F11	GND	H5	HA19
F12	V _{DD}	H6	TEST
F13	GND	H7	PSDCAS/PGPL3
F14	V _{DD}	H8	PGTA/PUPMWAIT/PGPL4/PPBS
F15	ETHRX_CLK/ETHSYNC_IN	H9	V _{DD}
F16	ETHTX_CLK/ETHREF_CLK/ETHCLOCK	H10	BM1/TC1/BNKSEL1
F17	GPIO20/TDM1RDAT	H11	ARTRY
F18	GPIO18/TDM1RSYN/DREQ2	H12	AACK
F19	GPIO16/TDM1TCLK/DONE1/DRACK1	H13	DBB/IRQ5
F20	GPIO11/TDM2TDAT/IRQ9/ETHRX_ER/ETHTXD	H14	HTA
F21	GPIO14/TDM2RDAT/IRQ12/ETHRXD0/NC	H15	V _{DD}
F22	GPIO19/TDM1RCLK/DACK2	H16	TT4/CS7
G2	HA24	H17	CS4
G3	HA27	H18	GPIO24/TDM0RSYN/IRQ14
G4	HA25	H19	GPIO21/TDM0TSYN
G5	HA23	H20	V _{DD}

Des.	Signal Name	Des.	Signal Name
H21	V _{DDH}	K15	V _{DD}
H22	A31	K16	TT2/CS5
J2	HA18	K17	ALE
J3	HA26	K18	CS2
J4	V _{DD}	K19	GND
J5	HA13	K20	A26
J6	GND	K21	A29
J7	PSDAMUX/PGPL5	K22	A28
J8	BADDR27	L2	HA12
J9	V _{DD}	L3	HA14
J10	CLKIN	L4	HA11
J11	BM2/TC2/BNKSEL2	L5	V _{DDH}
J12	DBG	L6	V _{DDH}
J13	V _{DD}	L7	BADDR28
J14	GND	L8	IRQ5/BADDR29
J15	V _{DD}	L9	GND
J16	TT3/CS6	L10	GND
J17	PSDA10/PGPL0	L14	GND
J18	BCTL1/CS5	L15	V _{DDH}
J19	GPIO23/TDM0TDAT/IRQ13	L16	GND
J20	GND	L17	GND
J21	GPIO25/TDM0RCLK/IRQ15	L18	CS3
J22	A30	L19	V _{DDH}
K2	HA15	L20	A27
K3	HA21	L21	A25
K4	HA16	L22	A22
K5	PWE3/PSDDQM3/PBS3	M2	HD28
K6	PWE1/PSDDQM1/PBS1	M3	HD31
K7	POE/PSDRAS/PGPL2	M4	V _{DDH}
K8	IRQ2/BADDR30	M5	GND
K9	Reserved	M6	GND
K10	GND	M7	GND
K11	GND	M8	V _{DD}
K12	GND	M9	V _{DDH}
K13	GND	M10	GND
K14	CLKOUT	M14	GND



Signal Name Signal Name Des. Des. M15 P12 V_{DDH} V_{CCSYN} M16 HBRST P13 GND M17 V_{DDH} P14 GND TA M18 P15 V_{DDH} BR GND M19 P16 TEA M20 V_{DDH} P17 PSDVAL P18 M21 A24 DP0/DREQ1/EXT_BR2 M22 A21 P19 N2 HD26 P20 V_{DDH} GND HD30 P21 N3 N4 HD29 P22 A19 N5 HD24 R2 HD18 PWE2/PSDDQM2/PBS2 N6 R3 V_{DDH} N7 VDDH R4 GND HWBS0/HDBS0/HWBE0/HDBE0 R5 HD22 N8 HBCS HWBS6/HDBS6/HWBE6/HDBE6/PWE6/PSDDQM6/PBS6 R6 N9 GND HWBS4/HDBS4/HWBE4/HDBE4/PWE4/PSDDQM4/PBS4 N10 R7 GND N14 R8 TSZ1 HRDS/HRW/HRDE N15 R9 TSZ3 BG IRQ1/GBL N16 R10 HCS N17 R11 V_{DD} N18 CS0 R12 V_{DD} PSDWE/PGPL1 N19 R13 V_{DD} N20 GPIO26/TDM0RDAT R14 TT0/HA7 IRQ7/DP7/DREQ4 N21 A23 R15 IRQ6/DP6/DREQ3 N22 A20 R16 IRQ3/DP3/DREQ2/EXT_BR3 P2 HD20 R17 TS P3 HD27 R18 IRQ2/DP2/DACK2/EXT_DBG2 P4 HD25 R19 Ρ5 HD23 R20 A17 HWBS3/HDBS3/HWBE3/HDBE3 P6 R21 A18 HWBS2/HDBS2/HWBE2/HDBE2 R22 A16 P7 HWBS1/HDBS1/HWBE1/HDBE1 T2 HD17 P8 Ρ9 HCLKIN HD21 T3 P10 GND Τ4 HD1/DSISYNC P11 **GND**_{SYN} T5 HD0/SWTE



Des.	Signal Name	Des.	Signal Name
T6	HWBS7/HDBS7/HWBE7/HDBE7/PWE7/PSDDQM7/PBS7	U21	A12
T7	HWBS5/HDBS5/HWBE5/HDBE5/PWE5/PSDDQM5/PBS5	U22	A13
Т8	TSZ0	V2	HD3/MODCK1
Т9	TSZ2	V3	V _{DDH}
T10	TBST	V4	GND
T11	V _{DD}	V5	D0
T12	D16	V6	D1
T13	TT1	V7	D4
T14	D21	V8	D5
T15	D23	V9	D7
T16	IRQ5/DP5/DACK4/EXT_BG3	V10	D10
T17	IRQ4/DP4/DACK3/EXT_DBG3	V11	D12
T18	IRQ1/DP1/DACK1/EXT_BG2	V12	D13
T19	D30	V13	D18
T20	GND	V14	D20
T21	A15	V15	GND
T22	A14	V16	D24
U2	HD16	V17	D27
U3	HD19	V18	D29
U4	HD2/DSI64	V19	A8
U5	D2	V20	A9
U6	D3	V21	A10
U7	D6	V22	A11
U8	D8	W2	HD6
U9	D9	W3	HD5/CNFGS
U10	D11	W4	HD4/MODCK2
U11	D14	W5	GND
U12	D15	W6	GND
U13	D17	W7	V _{DDH}
U14	D19	W8	V _{DDH}
U15	D22	W9	GND
U16	D25	W10	HDST1/HA10
U17	D26	W11	HDST0/HA9
U18	D28	W12	V _{DDH}
U19	D31	W13	GND
U20	V _{DDH}	W14	HD40/D40/ETHRXD0



Signal Name Signal Name Des. Des. W15 AA9 V_{DDH} V_{DDH} W16 HD33/D33/reserved AA10 HD54/D54/ETHTX_EN W17 VDDH AA11 HD52/D52 W18 HD32/D32/reserved AA12 V_{DDH} GND W19 GND AA13 W20 GND AA14 VDDH W21 A7 AA15 HD46/D46/ETHTXT0 W22 A6 AA16 GND HD42/D42/ETHRXD2/reserved Y2 HD7 AA17 HD15 AA18 Y3 HD38/D38/reserved Y4 AA19 HD35/D35/reserved V_{DDH} Y5 HD9 AA20 A0 Y6 V_DD AA21 A2 Y7 HD60/D60/ETHCOL/reserved AA22 A3 HD58/D58/ETHMDC GND Y8 AB2 Y9 GND AB3 HD13 Y10 AB4 HD11 V_{DDH} HD51/D51 Y11 AB5 HD8 Y12 GND AB6 HD62/D62 HD61/D61 Y13 AB7 VDDH HD43/D43/ETHRXD3/reserved Y14 AB8 HD57/D57/ETHRX_ER Y15 GND AB9 HD56/D56/ETHRX_DV/ETHCRS_DV Y16 V_{DDH} AB10 HD55/D55/ETHTX_ER/reserved Y17 GND AB11 HD53/D53 Y18 HD37/D37/reserved AB12 HD50/D50 Y19 HD34/D34/reserved AB13 HD49/D49/ETHTXD3/reserved Y20 AB14 HD48/D48/ETHTXD2/reserved V_{DDH} Y21 A4 AB15 HD47/D47/ETHTXD1 Y22 A5 AB16 HD45/D45 AA2 AB17 HD44/D44 V_{DD} AA3 HD14 AB18 HD41/D41/ETHRXD1 AA4 HD12 AB19 HD39/D39/reserved AA5 HD10 AB20 HD36/D36/reserved AA6 HD63/D63 AB21 A1 HD59/D59/ETHMDIO AB22 AA7 V_{DD} AA8 GND

Electrical Characteristics

2 Electrical Characteristics

This document contains detailed information on power considerations, DC/AC electrical characteristics, and AC timing specifications. For additional information, see the *MSC8122 Reference Manual*.

2.1 Maximum Ratings

CAUTION

This device contains circuitry protecting against damage due to high static voltage or electrical fields; however, normal precautions should be taken to avoid exceeding maximum voltage ratings. Reliability is enhanced if unused inputs are tied to an appropriate logic voltage level (for example, either GND or V_{DD}).

In calculating timing requirements, adding a maximum value of one specification to a minimum value of another specification does not yield a reasonable sum. A maximum specification is calculated using a worst case variation of process parameter values in one direction. The minimum specification is calculated using the worst case for the same parameters in the opposite direction. Therefore, a "maximum" value for a specification never occurs in the same device with a "minimum" value for another specification; adding a maximum to a minimum represents a condition that can never exist.

Table 2 describes the maximum electrical ratings for the MSC8122.

Table 2. Absolu	te Maximum	Ratings
-----------------	------------	---------

Rating	Symbol	Value	Unit
Core and PLL supply voltage	V _{DD}	-0.2 to 1.6	V
I/O supply voltage	V _{DDH}	-0.2 to 4.0	V
Input voltage	V _{IN}	-0.2 to 4.0	V
Maximum operating temperature: • Standard range • Extended range	TJ	90 105	ວ° ວ°
Minimum operating temperature • Standard range • Extended range	Τ _J	0 40	ů Č
Storage temperature range	T _{STG}	-55 to +150	°C

the listed limits may affect device reliability or cause permanent damage.

3. Section 3.5, Thermal Considerations includes a formula for computing the chip junction temperature (T_J).



rical Characteristics

2.2 Recommended Operating Conditions

Table 3 lists recommended operating conditions. Proper device operation outside of these conditions is not guaranteed.

Rating	Symbol	Value	Unit
Core and PLL supply voltage: • Standard — 400 MHz — 500 MHz • Reduced (300 and 400 MHz)	V _{DD} V _{CCSYN}	1.14 to 1.26 1.16 to 1.24 1.07 to 1.13	V V V
I/O supply voltage	V _{DDH}	3.135 to 3.465	V
Input voltage	V _{IN}	–0.2 to V _{DDH} +0.2	V
Operating temperature range: • Standard • Extended	T _J TJ	0 to 90 –40 to 105	℃ ℃

 Table 3. Recommended Operating Conditions

2.3 Thermal Characteristics

Table 4 describes thermal characteristics of the MSC8122 for the FC-PBGA packages.

	Observatoriatio	Symbol	FC- 20 ×	Unit	
	Characteristic		Natural Convection		200 ft/min (1 m/s) airflow
Junction-to-ambient ^{1, 2}		R _{θJA}	26	21	°C/W
Junction-to-ambient, four-layer board ^{1, 3}		R _{θJA}	19	15	°C/W
Junction-to-board (bottom) ⁴		R _{θJB}	9		°C/W
Junction-to-case ⁵		R _{θJC}	0.9		°C/W
Junction-to-package-top ⁶		Ψ_{JT}	1		°C/W
 Notes: 1. Junction temperature is a function of die size, on-chip power dissipation, package thermal resistance, mounting sit temperature, ambient temperature, air flow, power dissipation of other components on the board, and board therm resistance. 2. Per SEMI G38-87 and JEDEC JESD51-2 with the single layer board horizontal. 3. Per JEDEC JESD51-6 with the board horizontal. 4. Thermal resistance between the die and the printed circuit board per JEDEC JESD 51-8. Board temperature is me the top surface of the board near the package. 5. Thermal resistance between the die and the case top surface as measured by the cold plate method (MIL SPEC-8 1012.1). 					easured on

Thermal characterization parameter indicating the temperature difference between package top and the junction temperature

Table 4. Thermal Characteristics for the MSC8122

Section 3.5, *Thermal Considerations* provides a detailed explanation of these characteristics.

6.

per JEDEC JESD51-2.

Electrical Characteristics

DC Electrical Characteristics 2.4

This section describes the DC electrical characteristics for the MSC8122. The measurements in **Table 5** assume the following system conditions:

- ٠ $T_A = 25 \ ^\circ C$
- $V_{DD} =$ ٠
 - 300/400 MHz 1.1 V nominal = 1.07–1.13 V_{DC}
 - 400 MHz 1.2 V nominal = 1.14–1.26 V_{DC}
 - 500 MHz 1.2 V nominal = 1.16–1.24 V_{DC}
- $V_{DDH} = 3.3 \text{ V} \pm 5\% \text{ V}_{DC}$ ٠
- $GND = 0 V_{DC}$

Note: The leakage current is measured for nominal V_{DDH} and V_{DD} .

Characteristic	Symbol	Min	Typical	Max	Unit
Input high voltage ¹ , all inputs except CLKIN	V _{IH}	2.0	_	3.465	V
Input low voltage ¹	V _{IL}	GND	0	0.8	V
CLKIN input high voltage	V _{IHC}	2.4	3.0	3.465	V
CLKIN input low voltage	V _{ILC}	GND	0	0.8	V
Input leakage current, V _{IN} = V _{DDH}	I _{IN}	-1.0	0.09	1	μA
Tri-state (high impedance off state) leakage current, $V_{IN} = V_{DDH}$	I _{OZ}	-1.0	0.09	1	μA
Signal low input current, $V_{IL} = 0.8 V^2$	۱ _L	-1.0	0.09	1	μA
Signal high input current, $V_{IH} = 2.0 V^2$	I _H	-1.0	0.09	1	μA
Output high voltage, I _{OH} = -2 mA, except open drain pins	V _{OH}	2.0	3.0	—	V
Output low voltage, I _{OL} = 3.2 mA	V _{OL}	—	0	0.4	V
V _{CCSYN} PLL supply current	IVCCSYN	_	2	4	mA
Internal supply current: Wait mode Stop mode	I _{DDW} I _{DDS}		375 ³ 290 ³		mA mA
Typical power 400 MHz at 1.2 V ⁴	Р	_	1.15		W

e 5 for undershoot and overshoot voltages. igur

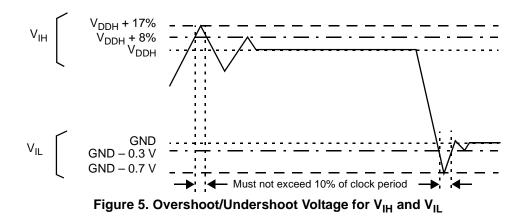
2. Not tested. Guaranteed by design.

Measured for 1.2 V core at 25°C junction temperature. 3.

The typical power values were measured using an EFR code with the device running at a junction temperature of 25°C. No 4. peripherals were enabled and the ICache was not enabled. The source code was optimized to use all the ALUs and AGUs and all four cores. It was created using CodeWarrior[®] 2.5. These values are provided as examples only. Power consumption is application dependent and varies widely. To assure proper board design with regard to thermal dissipation and maintaining proper operating temperatures, evaluate power consumption for your application and use the design guidelines in Chapter 4 of this document and in MSC8102, MSC8122, and MSC8126 Thermal Management Design Guidelines (AN2601).



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2.5 AC Timings

The following sections include illustrations and tables of clock diagrams, signals, and parallel I/O outputs and inputs. When systems such as DSP farms are developed using the DSI, use a device loading of 4 pF per pin. AC timings are based on a 20 pF load, except where noted otherwise, and a 50 Ω transmission line. For loads smaller than 20 pF, subtract 0.06 ns per pF down to 10 pF load. For loads larger than 20 pF, add 0.06 ns for SIU/Ethernet/DSI delay and 0.07 ns for GPIO/TDM/timer delay. When calculating overall loading, also consider additional RC delay.

2.5.1 Output Buffer Impedances

Table 6. Output Buffer Impedances

Output Buffers	Typical Impedance (Ω)		
System bus	50		
Memory controller	50		
Parallel I/O	50		
Note: These are typical values at 65°C. The impedance may vary by ±25% depending on device process and operating temperature.			

2.5.2 Start-Up Timing

Starting the device requires coordination among several input sequences including clocking, reset, and power. **Section 2.5.3** describes the clocking characteristics. **Section 2.5.4** describes the reset and power-up characteristics. You must use the following guidelines when starting up an MSC8122 device:

- **PORESET** and **TRST** must be asserted externally for the duration of the power-up sequence. See **Table 11** for timing.
- If possible, bring up the V_{DD} and V_{DDH} levels together. For designs with separate power supplies, bring up the V_{DD} levels and then the V_{DDH} levels (see **Figure 7**).
- CLKIN should start toggling at least 16 cycles (starting after V_{DDH} reaches its nominal level) before PORESET deassertion to guarantee correct device operation (see Figure 6 and Figure 7).
- CLKIN must not be pulled high during V_{DDH} power-up. CLKIN can toggle during this period.
- **Note:** See Section 3.1 for start-up sequencing recommendations and Section 3.2 for power supply design recommendations.

The following figures show acceptable start-up sequence examples. Figure 6 shows a sequence in which V_{DD} and V_{DDH} are raised together. Figure 7 shows a sequence in which V_{DDH} is raised after V_{DD} and CLKIN begins to toggle as V_{DDH} rises.



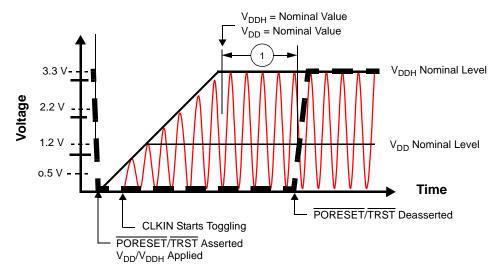


Figure 6. Start-Up Sequence: V_{DD} and V_{DDH} Raised Together

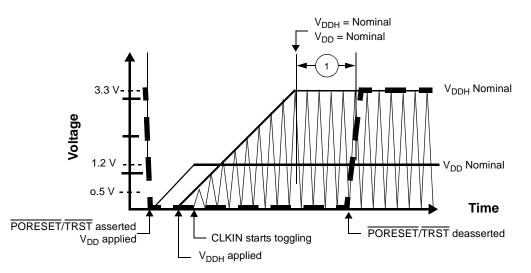


Figure 7. Start-Up Sequence: V_{DD} Raised Before V_{DDH} with CLKIN Started with V_{DDH}



In all cases, the power-up sequence must follow the guidelines shown in Figure 8.

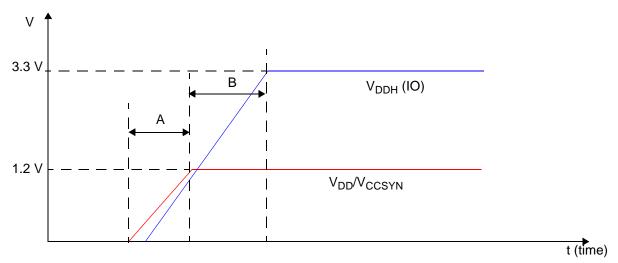


Figure 8. Power-Up Sequence for V_{DDH} and V_{DD}/V_{CCSYN}

The following rules apply:

- 1. During time interval A, V_{DDH} should always be equal to or less than the V_{DD}/V_{CCSYN} voltage level. The duration of interval A should be kept below 10 ms.
- 2. The duration of timing interval B should be kept as small as possible and less than 10 ms.

2.5.3 Clock and Timing Signals

The following sections include a description of clock signal characteristics. **Table 7** shows the maximum frequency values for internal (Core, Reference, Bus, and DSI) and external (CLKIN and CLKOUT) clocks. The user must ensure that maximum frequency values are not exceeded.

Table 7.	Maximum	Frequencies
----------	---------	-------------

Characteristic	Maximum in MHz
Core frequency	300/400/500
Reference frequency (REFCLK)	100/133/166
Internal bus frequency (BLCK)	100/133/166
DSI clock frequency (HCLKIN)	
Core frequency = 300 MHz	$HCLKIN \leq (min\{70 MHz, CLKOUT\})$
Core frequency = 400/500 MHz	HCLKIN ≤ (min{100 MHz, CLKOUT})
External clock frequency (CLKIN or CLKOUT)	100/133/166

Table 8.	Clock	Frequencies
----------	-------	-------------

Characteristics	0h.al	300 MHz Device		400 MHz Device		500 MHz Device		
	Symbol	Min	Мах	Min	Мах	Min	Max	
CLKIN frequency	F _{CLKIN}	20	100	20	133.3	20	166.7	
BCLK frequency	F _{BCLK}	40	100	40	133.3	40	166.7	
Reference clock (REFCLK) frequency	F _{REFCLK}	40	100	40	133.3	40	166.7	
Output clock (CLKOUT) frequency	F _{CLKOUT}	40	100	40	133.3	40	166.7	
SC140 core clock frequency	F _{CORE}	200	300	200	400	200	500	
Note: The rise and fall time of external clocks should be 3 ns maximum								



Characteristic	Min	Max	Unit
Phase jitter between BCLK and CLKIN	—	0.3	ns
CLKIN frequency	20	see Table 8	MHz
CLKIN slope	—	3	ns
CLKIN period jitter ¹	—	150	ps
CLKIN jitter spectrum	150	—	KHz
PLL input clock (after predivider)	20	100	MHz
PLL output frequency (VCO output)	800		MHz
300 MHz core		1200	MHz
400 MHz core		1600	MHz
500 MHz core		2000	MHz
CLKOUT frequency jitter ¹	—	200	ps
CLKOUT phase jitter ¹ with CLKIN phase jitter of ±100 ps.	—	500	ps
Notes:1.Peak-to-peak.2.Not tested. Guaranteed by design.			

2.5.4 Reset Timing

The MSC8122 has several inputs to the reset logic:

- Power-on reset (PORESET)
- External hard reset (HRESET)
- External soft reset (SRESET)
- Software watchdog reset
- Bus monitor reset
- Host reset command through JTAG

All MSC8122 reset sources are fed into the reset controller, which takes different actions depending on the source of the reset. The reset status register indicates the most recent sources to cause a reset. **Table 10** describes the reset sources.

Table 10. Reset Sources

Name	Direction	Description
Power-on reset (PORESET)	Input	Initiates the power-on reset flow that resets the MSC8122 and configures various attributes of the MSC8122. On PORESET, the entire MSC8122 device is reset. SPLL states is reset, HRESET and SRESET are driven, the SC140 extended cores are reset, and system configuration is sampled. The clock mode (MODCK bits), reset configuration mode, boot mode, Chip ID, and use of either a DSI 64 bits port or a System Bus 64 bits port are configured only when PORESET is asserted.
External hard reset (HRESET)	Input/ Output	Initiates the hard reset flow that configures various attributes of the MSC8122. While HRESET is asserted, SRESET is also asserted. HRESET is an open-drain pin. Upon hard reset, HRESET and SRESET are driven, the SC140 extended cores are reset, and system configuration is sampled. The most configurable features are reconfigured. These features are defined in the 32-bit hard reset configuration word described in <i>Hard Reset Configuration Word</i> section of the <i>Reset</i> chapter in the <i>MSC8122 Reference Manual</i> .
External soft reset (SRESET)	Input/ Output	Initiates the soft reset flow. The MSC8122 detects an external assertion of SRESET only if it occurs while the MSC8122 is not asserting reset. SRESET is an open-drain pin. Upon soft reset, SRESET is driven, the SC140 extended cores are reset, and system configuration is maintained.
Software watchdog reset	Internal	When the MSC8122 watchdog count reaches zero, a software watchdog reset is signalled. The enabled software watchdog event then generates an internal hard reset sequence.
Bus monitor reset	Internal	When the MSC8122 bus monitor count reaches zero, a bus monitor hard reset is asserted. The enabled bus monitor event then generates an internal hard reset sequence.
Host reset command through the TAP	Internal	When a host reset command is written through the Test Access Port (TAP), the TAP logic asserts the soft reset signal and an internal soft reset sequence is generated.

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Table 11 summarizes the reset actions that occur as a result of the different reset sources.

Reset Action/Reset Source	Power-On <u>Reset</u> (PORESET)	Hard Reset (HRESET) Soft Reset		Reset (SRESET)
Reset Action/Reset Source	External only	External or Internal (Software Watchdog or Bus Monitor)	External	JTAG Command: EXTEST, CLAMP, or HIGHZ
Configuration pins sampled (Refer to Section 2.5.4.1 for details).	Yes	No	No	No
SPLL state reset	Yes	No	No	No
System reset configuration write through the DSI	Yes	No	No	No
System reset configuration write though the system bus	Yes	Yes	No	No
HRESET driven	Yes	Yes	No	No
SIU registers reset	Yes	Yes	No	No
IPBus modules reset (TDM, UART, Timers, DSI, IPBus master, GIC, HS, and GPIO)	Yes	Yes	Yes	Yes
SRESET driven	Yes	Yes	Yes	Depends on command
SC140 extended cores reset	Yes	Yes	Yes	Yes
MQBS reset	Yes	Yes	Yes	Yes

Table 11. Reset Actions for Each Reset Source

2.5.4.1 Power-On Reset (PORESET) Pin

Asserting $\overrightarrow{PORESET}$ initiates the power-on reset flow. $\overrightarrow{PORESET}$ must be asserted externally for at least 16 CLKIN cycles after V_{DD} and V_{DDH} are both at their nominal levels.

2.5.4.2 Reset Configuration

The MSC8122 has two mechanisms for writing the reset configuration:

- Through the direct slave interface (DSI)
- Through the system bus. When the reset configuration is written through the system bus, the MSC8122 acts as a configuration master or a configuration slave. If configuration slave is selected, but no special configuration word is written, a default configuration word is applied.

Fourteen signal levels (see **Chapter 1** for signal description details) are sampled on **PORESET** deassertion to define the Reset Configuration Mode and boot and operating conditions:

- RSTCONF
- CNFGS
- DSISYNC
- DSI64
- CHIP_ID[0-3]
- BM[0–2]
- SWTE
- MODCK[1-2]

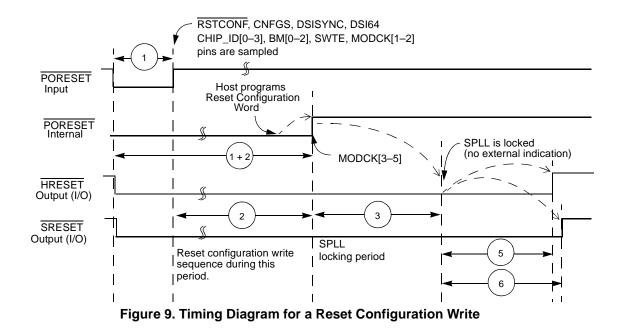


2.5.4.3 Reset Timing Tables

Table 12 and Figure 9 describe the reset timing for a reset configuration write through the direct slave interface (DSI) or through the system bus.

No.	Characteristics	Expression	Min	Max	Unit
1	Required external PORESET duration minimum • CLKIN = 20 MHz • CLKIN = 100 MHz (300 MHz core) • CLKIN = 133 MHz (400 MHz core) • CLKIN = 166 MHz (500 MHz core)	16/CLKIN	800 160 120 96		ns ns ns ns
2	Delay from deassertion of external PORESET to deassertion of internal PORESET • CLKIN = 20 MHz to 166 MHz	1024/CLKIN	6.17	51.2	μs
3	Delay from de-assertion of internal PORESET to SPLL lock CLKIN = 20 MHz (RDF = 1) CLKIN = 100 MHz (RDF = 1) (300 MHz core) CLKIN = 133 MHz (RDF = 2) (400 MHz core) CLKIN = 166 MHz (RDF = 2) (500 MHz core) 	6400/(CLKIN/RDF) (PLL reference clock-division factor)	320 64 96 77	320 64 96 77	hs hs hs
5	Delay from SPLL to HRESET deassertion REFCLK = 40 MHz to 166 MHz 	512/REFCLK	3.08	12.8	μs
6	Delay from SPLL lock to SRESET deassertion REFCLK = 40 MHz to 166 MHz 	515/REFCLK	3.10	12.88	μs
7	Setup time from assertion of RSTCONF, CNFGS, DSISYNC, DSI64, CHIP_ID[0–3], BM[0–2], SWTE, and MODCK[1–2] before deassertion of PORESET		3	_	ns
8	Hold time from deassertion of PORESET to deassertion of RSTCONF, CNFGS, DSISYNC, DSI64, CHIP_ID[0–3], BM[0–2], SWTE, and MODCK[1–2]		5	—	ns
Note:	Timings are not tested, but are guaranteed by design.		•	•	•

 Table 12. Timing for a Reset Configuration Write through the DSI or System Bus



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System Bus Access Timing 2.5.5

Core Data Transfers 2.5.5.1

Generally, all MSC8122 bus and system output signals are driven from the rising edge of the reference clock (REFCLK). The REFCLK is the CLKIN signal. Memory controller signals, however, trigger on four points within a REFCLK cycle. Each cycle is divided by four internal ticks: T1, T2, T3, and T4. T1 always occurs at the rising edge of REFCLK (and T3 at the falling edge), but the spacing of T2 and T4 depends on the PLL clock ratio selected, as Table 13 shows.

BCLK/SC140 clock	Tick Spacing (T1 Occurs at the Rising Edge of REFCLK)					
	T2	Т3	T4			
1:4, 1:6, 1:8, 1:10	1/4 REFCLK	1/2 REFCLK	3/4 REFCLK			
1:3	1/6 REFCLK	1/2 REFCLK	4/6 REFCLK			
1:5	2/10 REFCLK	1/2 REFCLK	7/10 REFCLK			

Table 13. Tick Spacing for Memory Controller Signals

Figure 10 is a graphical representation of Table 13.

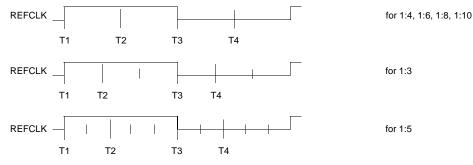


Figure 10. Internal Tick Spacing for Memory Controller Signals



The UPM machine and GPCM machine outputs change on the internal tick selected by the memory controller configuration. The AC timing specifications are relative to the internal tick. SDRAM machine outputs change only on the REFCLK rising edge.

		Value for Bus Speed in MHz				
		R	ef = CLK	IN	Ref = CLKOUT	
No.	Characteristic		1.2 V	1.2 V	1.2 V	Units
		100/ 133	133	166	133	
10	Hold time for all signals after the 50% level of the REFCLK rising edge	0.5	0.5	0.5	0.5	ns
11a	ARTRY/ABB set-up time before the 50% level of the REFCLK rising edge	3.1	3.0	3.0	3.0	ns
11b	DBG/DBB/BG/BR/TC set-up time before the 50% level of the REFCLK rising edge	3.6	3.3	3.3	3.3	ns
11c	AACK set-up time before the 50% level of the REFCLK rising edge	3.0	2.9	2.9	2.9	ns
11d	TA/TEA/PSDVAL set-up time before the 50% level of the REFCLK rising edge					
	Data-pipeline modeNon-pipeline mode	3.5 4.4	3.4 4.0	3.4 4.0	3.4 4.0	ns ns
12	Data bus set-up time before REFCLK rising edge in Normal mode Data-pipeline mode Non-pipeline mode 	1.9 4.2	1.8 4.0	1.7 4.0	1.8 4.0	ns ns
13 ¹	Data bus set-up time before the 50% level of the REFCLK rising edge in ECC and PARITY modes • Data-pipeline mode • Non-pipeline mode	2.0 8.2	2.0 7.3	2.0 7.3	2.0 7.3	ns ns
14 ¹	DP set-up time before the 50% level of the REFCLK rising edge Data-pipeline mode Non-pipeline mode 	2.0 7.9	2.0 6.1	2.0 6.1	2.0 6.1	ns ns
15a	 TS and Address bus set-up time before the 50% level of the REFCLK rising edge Extra cycle mode (SIUBCR[EXDD] = 0) No extra cycle mode (SIUBCR[EXDD] = 1) 	4.2 5.5	3.8 5.0	3.8 5.0	3.8 5.0	ns ns
15b	Address attributes: TT/TBST/TSZ/GBL set-up time before the 50% level of the REFCLK rising edge • Extra cycle mode (SIUBCR[EXDD] = 0) • No extra cycle mode (SIUBCR[EXDD] = 1)	3.7 4.8	3.5 4.4	3.5 4.4	3.5 4.4	ns ns
16	PUPMWAIT signal set-up time before the 50% level of the REFCLK rising edge	3.7	3.7	3.7	3.7	ns
17	IRQx setup time before the 50% level; of the REFCLK rising edge ³	4.0	4.0	4.0	4.0	ns
18	IRQx minimum pulse width ³	6.0 + T _{REFCLK}	6.0 + T _{REFCLK}	6.0 + T _{REFCLK}	6.0 + T _{REFCLK}	ns
Notes:	 Timings specifications 13 and 14 in non-pipeline mode are more restrictive than MSC8102 timings. Values are measured from the 50% TTL transition level relative to the 50% level of the REFCLK rising edge. Guaranteed by design. 					

Table 14. AC Timing for SIU Inputs

		V	alue for	Bus Spe	ed in MHz ³	
		R	ef = CLK	IN	Ref = CLKOUT	
No.	Characteristic	1.1 V	1.2 V	1.2 V	1.2 V	Units
		100/ 133	133	166	100/133	
30 ²	Minimum delay from the 50% level of the REFCLK for all signals	0.9	0.8	0.8	1.0	ns
31	PSDVAL/TEA/TA max delay from the 50% level of the REFCLK rising edge	6.0	4.9	4.9	5.8	ns
32a	 Address bus max delay from the 50% level of the REFCLK rising edge Multi-master mode (SIUBCR[EBM] = 1) Single-master mode (SIUBCR[EBM] = 0) 	6.4 5.3	5.5 4.2	5.5 3.9	6.4 5.1	ns ns
32b	Address attributes: TT[0–1]/TBST/TSZ/GBL max delay from the 50% level of the REFCLK rising edge	6.4	5.1	5.1	6.0	ns
32c	Address attributes: TT[2–4]/TC max delay from the 50% level of the REFCLK rising edge	6.9	5.7	5.7	6.6	ns
32d	BADDR max delay from the 50% level of the REFCLK rising edge	5.2	4.2	4.2	5.1	ns
33a	Data bus max delay from the 50% level of the REFCLK rising edgeData-pipeline modeNon-pipeline mode	4.8 7.1	3.9 6.1	3.7 6.1	4.8 7.0	ns ns
33b	DP max delay from the 50% level of the REFCLK rising edgeData-pipeline modeNon-pipeline mode	6.0 7.5	5.3 6.5	5.3 6.5	6.2 7.4	ns ns
34	Memory controller signals/ALE/CS[0–4] max delay from the 50% level of the REFCLK rising edge	5.1	4.2	3.9	5.1	ns
35a	DBG/BG/BR/DBB max delay from the 50% level of the REFCLK rising edge	6.0	4.7	4.7	5.6	ns
35b	AACK/ABB/TS/CS[5–7] max delay from the 50% level of the REFCLK rising edge	5.5	4.5	4.5	5.4	ns

Decreasing the load results in a timing decrease at the rate of 0.3 ns per 5 pF decrease in load. Increasing the load results in

• In 60x-compatible mode connected to another MSC8122 device, the frequency is determined by adding the input and output longest timing values, which results in the total delay for 20 pF output capacitance. You must also account for other

• To achieve maximum performance on the bus in single-master mode, disable the DBB signal by writing a 1 to the

influences that can affect timing, such as on-board clock skews, on-board noise delays, and so on.
In single-master mode, the frequency depends on the timing of the devices connected to the MSC8122.

SIUMCR[BDD] bit. See the SIU chapter in the MSC8122 Reference Manual for details.

a timing increase at the rate of 0.15 ns per 5 pF increase in load.

The maximum bus frequency depends on the mode:

Table 15. AC Timing for SIU Outputs

3.

Electrical Characteristics



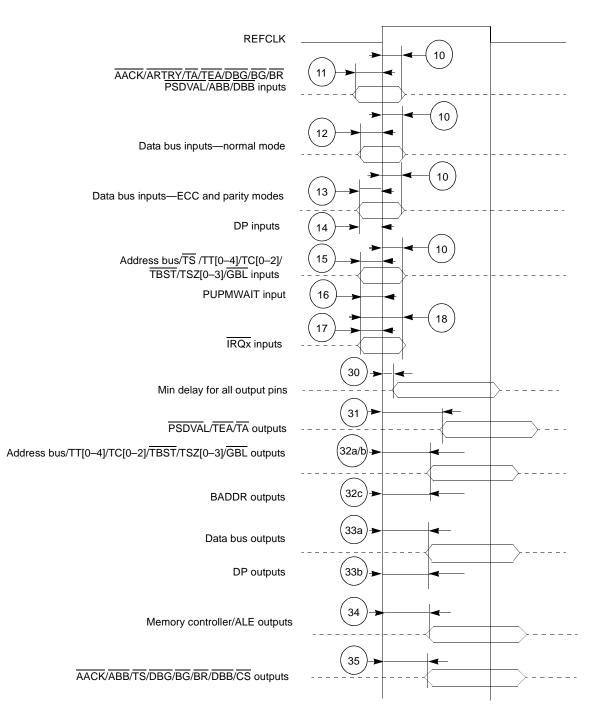


Figure 11. SIU Timing Diagram



rical Characteristics

2.5.5.2 CLKIN to CLKOUT Skew

Table 17 describes the CLKOUT-to-CLKIN skew timing.

No.	Characteristic	Min ¹	Max ¹	Units
20	Rise-to-rise skew			
	• V _{DD} = 1.1 V	0.0	0.95	ns
	• $V_{DD} = 1.2 V$	0.0	0.85	ns
21	Fall-to-fall skew			
	• V _{DD} = 1.1 V	-1.5	1.0	ns
	• V _{DD} = 1.2 V	-0.8	1.0	ns
22	CLKOUT phase (1.2 V, 133 MHz)			
	Phase high	2.8	_	ns
	Phase low	2.8	—	ns
23	CLKOUT phase (1.1 V, 133 MHz)			
	Phase high	2.2	—	ns
	Phase low	2.2	—	ns
24	CLKOUT phase (1.1 V, 100 MHz)			
	Phase high	3.3	_	ns
	Phase low	3.3	—	ns
Notes:	1. A positive number indicates that CLKOUT precedes CLKIN, A negative nur	mber indicates that C	LKOUT follows CLK	IN.
	2. Skews are measured in clock mode 29, with a CLKIN:CLKOUT ratio of 1:1.	The same skew is v	alid for all clock mo	des.
	3. CLKOUT skews are measured using a load of 10 pF.			
	4. CLKOUT skews and phase are not measured for 500/166 Mhz parts becau	se these parts only	use CLKIN mode.	

For designs that use the CLKOUT synchronization mode, use the skew values listed in **Table 16** to adjust the rise-to-fall timing values specified for CLKIN synchronization. **Figure 12** shows the relationship between the CLKOUT and CLKIN timings.



Figure 12. CLKOUT and CLKIN Signals.



2.5.5.3 DMA Data Transfers

Table 17 describes the DMA signal timing.

No.	Characteristic	Ref =	CLKIN	Ref = C (1.2 V	Units	
		Min	Max	Min	Max	
37	DREQ set-up time before the 50% level of the falling edge of REFCLK	5.0	—	5.0	—	ns
38	DREQ hold time after the 50% level of the falling edge of REFCLK	0.5	—	0.5	—	ns
39	DONE set-up time before the 50% level of the rising edge of REFCLK	5.0	_	5.0	—	ns
40	DONE hold time after the 50% level of the rising edge of REFCLK	0.5	_	0.5	—	ns
41	DACK/DRACK/DONE delay after the 50% level of the REFCLK rising edge	0.5	7.5	0.5	8.4	ns

Table 17. DMA Signals

The DREQ signal is synchronized with REFCLK. To achieve fast response, a synchronized peripheral should assert DREQ according to the timings in **Table 17**. Figure 13 shows synchronous peripheral interaction.

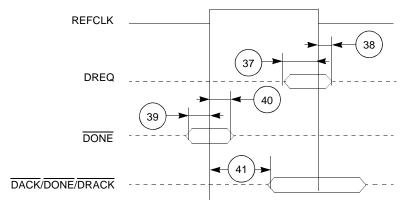


Figure 13. DMA Signals



2.5.6 DSI Timing

The timings in the following sections are based on a 20 pF capacitive load.

2.5.6.1 DSI Asynchronous Mode

Table 1	8.	DSI	Asv	vnchr	onous	Mode	Timing
	υ.	201	A 3)	y 110111	Unous	mouc	i iiiiiiig

No.	Characteristics	Min	Max	Unit
100	Attributes ¹ set-up time before strobe (HWBS[n]) assertion	1.5	—	ns
101	Attributes ¹ hold time after data strobe deassertion	1.3	—	ns
102	Read/Write data strobe deassertion width: • DCR[HTAAD] = 1			
	— Consecutive access to the same DSI — Different device with DCR[HTADT] = 01	1.8 + T _{REFCLK} 5 + T _{REFCLK}		ns ns
	 Different device with DCR[HTADT] = 10 Different device with DCR[HTADT] = 11 	$5 + (1.5 \times T_{REFCLK})$ $5 + (2.5 \times T_{REFCLK})$		ns ns
	 DCR[HTAAD] = 0 	$1.8 + T_{REFCLK}$		ns
103	Read data strobe deassertion to output data high impedance	—	8.5	ns
104	Read data strobe assertion to output data active from high impedance	2.0	_	ns
105	Output data hold time after read data strobe deassertion	2.2	_	ns
106	Read/Write data strobe assertion to HTA active from high impedance	2.2	_	ns
107	Output data valid to HTA assertion	3.2	—	ns
108	Read/Write data strobe assertion to HTA valid ² 1.1 V core 	_	7.4	ns
	• 1.2 V core	_	6.7	ns
109	Read/Write data strobe deassertion to output \overline{HTA} high impedance. (DCR[HTAAD] = 0, \overline{HTA} at end of access released at logic 0)	-	6.5	ns
110	Read/Write data strobe deassertion to output HTA deassertion. (DCR[HTAAD] = 1, HTA at end of access released at logic 1)	-	6.5	ns
111	Read/Write data strobe deassertion to output HTA high impedance. (DCR[HTAAD] = 1, HTA at end of access released at logic 1	_		
	• DCR[HTADT] = 01		5 + T _{REFCLK}	ns
	DCR[HTADT] = 10		$5 + (1.5 \times T_{REFCLK})$	ns
	DCR[HTADT] = 11		5 + (2.5 \times T _{REFCLK})	ns
112	Read/Write data strobe assertion width	1.8 + T _{REFCLK}	_	ns
201	Host data input set-up time before write data strobe deassertion	1.0	—	ns
202	Host data input hold time after write data strobe deassertion	4 7		
	1.1 V core 1.2 V core	1.7 1.5		ns ns
Notes:	 Attributes refers to the following signals: HCS, HA[11–29], HCID[0– 	_	and UWRSn	113
Notes.	 Autobaces refers to the following signals. Hoo, HA[11, 25], Hop[0] This specification is tested in dual-strobe mode. Timing in single-str All values listed in this table are tested or guaranteed by design. 	•		

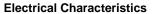
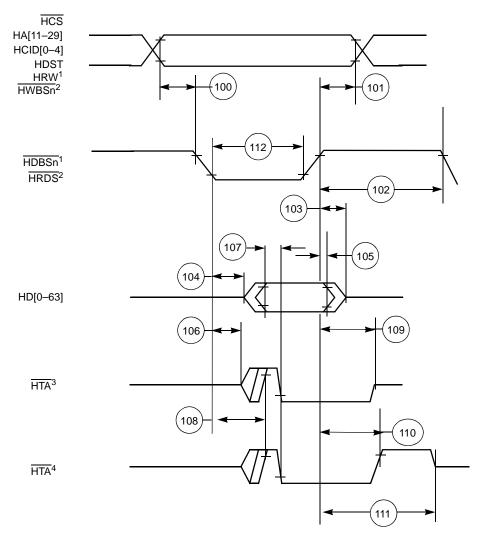


Figure 14 shows DSI asynchronous read signals timing.



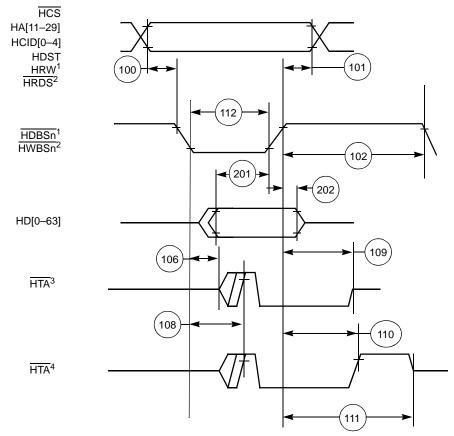
Notes: 1. Used for single-strobe mode access.

- **2.** Used for dual-strobe mode access.
- **3.** HTA released at logic 0 (DCR[HTAAD] = 0) at end of access; used with pull-down implementation.
- 4. HTA released at logic 1 (DCR[HTAAD] = 1) at end of access; used with pull-up implementation.

Figure 14. Asynchronous Single- and Dual-Strobe Modes Read Timing Diagram



Figure 15 shows DSI asynchronous write signals timing.



Notes: 1. Used for single-strobe mode access.

- 2. Used for dual-strobe mode access.
- 3. HTA released at logic 0 (DCR[HTAAD] = 0) at end of access; used with pull-down implementation.
- 4. HTA released at logic 1 (DCR[HTAAD] = 1) at end of access; used with pull-up implementation.

Figure 15. Asynchronous Single- and Dual-Strobe Modes Write Timing Diagram

Figure 16 shows DSI asynchronous broadcast write signals timing.

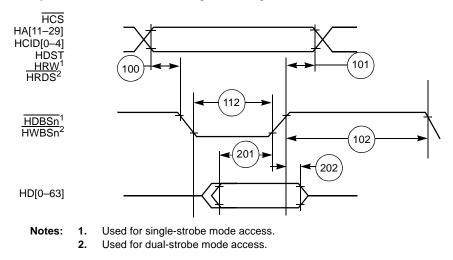


Figure 16. Asynchronous Broadcast Write Timing Diagram

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2.5.6.2 DSI Synchronous Mode

Table 19. DSI Inputs in Synchronous Mode

No.	Characteristic	Expression -	1.1 V	Core	1.2 V	Units	
NO.	Characteristic		Min	Max	Min	Max	Units
120	HCLKIN cycle time ^{1,2}	HTC	10.0	55.6	10.0	55.6	ns
121	HCLKIN high pulse width	$(0.5\pm0.1)\times HTC$	4.0	33.3	4.0	33.3	ns
122	HCLKIN low pulse width	$(0.5\pm0.1)\times HTC$	4.0	33.3	4.0	33.3	ns
123	HA[11–29] inputs set-up time	—	1.2	—	1.2	—	ns
124	HD[0–63] inputs set-up time	—	0.6	—	0.4	—	ns
125	HCID[0-4] inputs set-up time	—	1.3	—	1.3	—	ns
126	All other inputs set-up time	—	1.2	—	1.2	—	ns
127	All inputs hold time	—	1.5	—	1.5	_	ns
Notes:	 Values are based on a frequency range of 18–100 MH Refer to Table 7 for HCLKIN frequency limits. 	Z.					

Table 20. DSI Outputs in Synchronous Mode

No.	Characteristic	1.1 V	Core	1.2 V	Units	
NO.	Characteristic	Min	Max	Min	Max	Units
128	HCLKIN high to HD[0–63] output active	2.0	_	2.0	_	ns
129	HCLKIN high to HD[0–63] output valid		7.6	_	6.3	ns
130	HD[0–63] output hold time	1.7	—	1.7	—	ns
131	HCLKIN high to HD[0–63] output high impedance	_	8.3	_	7.6	ns
132	HCLKIN high to HTA output active	2.2	_	2.0	—	ns
133	HCLKIN high to HTA output valid	_	7.4	_	5.9	ns
134	HTA output hold time	1.7	_	1.7	—	ns
135	HCLKIN high to HTA high impedance	_	7.5	—	6.3	ns

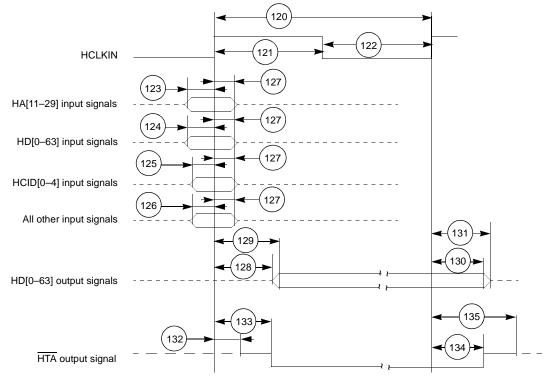


Figure 17. DSI Synchronous Mode Signals Timing Diagram





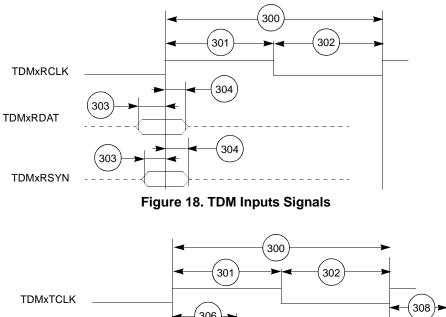
2.5.7 TDM Timing

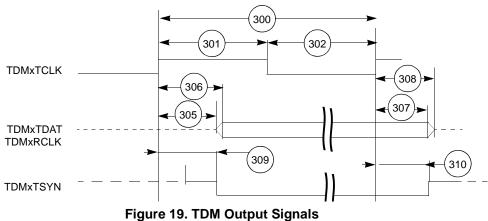
Na	Characteristic	Evenesion	1.1 V	Core	1.2 V Core		Unito
No.		Expression	Min	Max	Min	Max	Units
300	TDMxRCLK/TDMxTCLK	TC ¹	16	—	16	_	ns
301	TDMxRCLK/TDMxTCLK high pulse width	$(0.5\pm0.1) imes TC$	7	—	7	—	ns
302	TDMxRCLK/TDMxTCLK low pulse width	$(0.5\pm0.1) imes TC$	7	—	7	_	ns
303	TDM receive all input set-up time		1.3	—	1.3	_	ns
304	TDM receive all input hold time		1.0	—	1.0	_	ns
305	TDMxTCLK high to TDMxTDAT/TDMxRCLK output active ^{2,3}		2.8	_	2.8	—	ns
306	TDMxTCLK high to TDMxTDAT/TDMxRCLK output		_	10.0	—	8.8	ns
307	All output hold time ⁴		2.5	—	2.5	_	ns
308	TDMxTCLK high to TDmXTDAT/TDMxRCLK output high impedance ^{2,3}		—	10.7	—	10.5	ns
309	TDMxTCLK high to TDMXTSYN output valid ²		—	9.7	—	8.5	ns
310	TDMxTSYN output hold time ⁴		2.5	—	2.5	—	ns
Notes:	 Values are based on a a maximum frequency of 62.5 Devices operating at 300 MHz are limited to a maximu Values are based on 20 pF capacitive load. 					low 62.5 l	MHz.

Table 21. TDM Timing

3. When configured as an output, TDMxRCLK acts as a second data link. See the MSC8122 Reference Manual for details.

4. Values are based on 10 pF capacitive load.







2.5.8 UART Timing

No.	Characteristics	Expression	Min	Max	Unit
400	URXD and UTXD inputs high/low duration	16 × T _{REFCLK}	160.0	—	ns
401	URXD and UTXD inputs rise/fall time			10	ns
402	UTXD output rise/fall time			10	ns



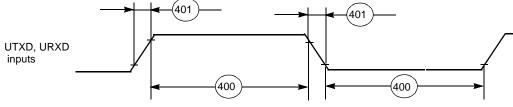


Figure 20. UART Input Timing

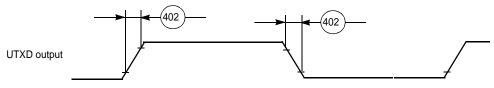


Figure 21. UART Output Timing

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2.5.9 Timer Timing

No.	Characteristics	Ref =	Unit	
NO.		Min	Max	Unit
500	TIMERx frequency	10.0	_	ns
501	TIMERx Input high period	4.0		ns
502	TIMERx Output low period	4.0		ns
503	 TIMERx Propagations delay from its clock input 1.1 V core 1.2 V core 	3.1 2.8	9.5 8.1	ns ns



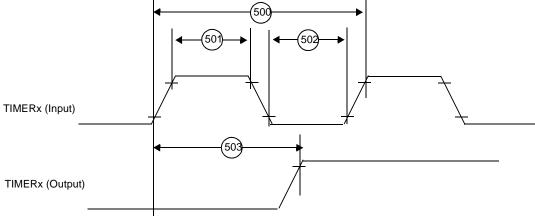


Figure 22. Timer Timing

2.5.10 Ethernet Timing

2.5.10.1 Management Interface Timing

Table 24. Ethernet Controller Management Interface Timing

No.	Characteristics	Min	Max	Unit
801	ETHMDIO to ETHMDC rising edge set-up time	10	_	ns
802	ETHMDC rising edge to ETHMDIO hold time	10		ns

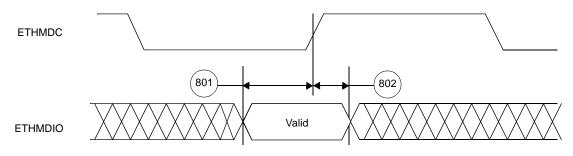


Figure 23. MDIO Timing Relationship to MDC



2.5.10.2 MII Mode Timing

No.	Characteristics	Min	Max	Unit
803	ETHRX_DV, ETHRXD[0–3], ETHRX_ER to ETHRX_CLK rising edge set-up time	3.5	—	ns
804	ETHRX_CLK rising edge to ETHRX_DV, ETHRXD[0–3], ETHRX_ER hold time	3.5	—	ns
805	ETHTX_CLK to ETHTX_EN, ETHTXD[0–3], ETHTX_ER output delay			
	• 1.1 V core	1	14.6	ns
	• 1.2 V core	1	12.6	ns



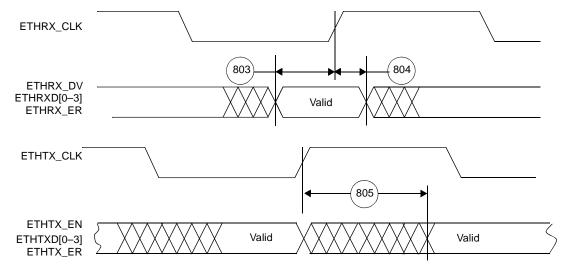


Figure 24. MII Mode Signal Timing

2.5.10.3 RMII Mode

No.	Characteristics	1.1 V Core		1.2 V Core		Unit
		Min	Max	Min	Max	Omt
806	ETHTX_EN,ETHRXD[0–1], ETHCRS_DV, ETHRX_ER to ETHREF_CLK rising edge set-up time	1.6		2	—	ns
807	ETHREF_CLK rising edge to ETHRXD[0–1], ETHCRS_DV, ETHRX_ER hold time	1.6	_	1.6	_	ns
811	ETHREF_CLK rising edge to ETHTXD[0–1], ETHTX_EN output delay.	3	12.5	3	11	ns

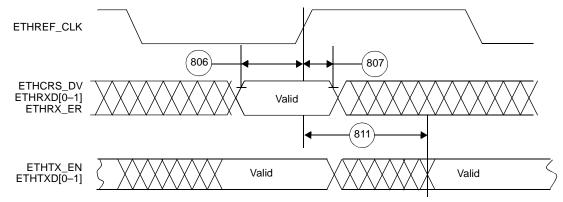


Figure 25. RMII Mode Signal Timing

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2.5.10.4 SMII Mode

No.	Characteristics	Min	Max	Unit
808	ETHSYNC_IN, ETHRXD to ETHCLOCK rising edge set-up time	1.0	_	ns
809	ETHCLOCK rising edge to ETHSYNC_IN, ETHRXD hold time	1.0	_	ns
810	ETHCLOCK rising edge to ETHSYNC, ETHTXD output delay1.1 V core.1.2 V core.	1.5 ¹ 1.5 ¹	6.0 ² 5.0 ²	ns ns
Notes:	 Measured using a 5 pF load. Measured using a 15 pF load. 			



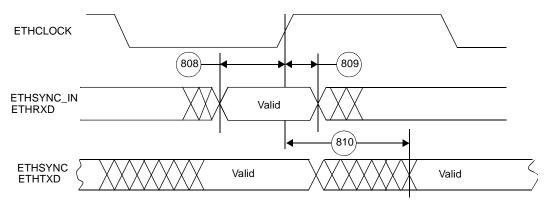


Figure 26. SMII Mode Signal Timing

2.5.11 GPIO Timing

Table 28. GPIO Timing

No.	Characteristics	Ref = CLKIN		Ref = CLKOUT (1.2 V only)		Unit
		Min	Max	Min	Max	1
601	REFCLK edge to GPIO out valid (GPIO out delay time)	—	6.1	_	6.9	ns
602	REFCLK edge to GPIO out not valid (GPIO out hold time)	1.1		1.3	—	ns
603	REFCLK edge to high impedance on GPIO out	—	5.4		6.2	ns
604	GPIO in valid to REFCLK edge (GPIO in set-up time)	3.5	_	3.7	_	ns
605	REFCLK edge to GPIO in not valid (GPIO in hold time)	0.5		0.5	_	ns



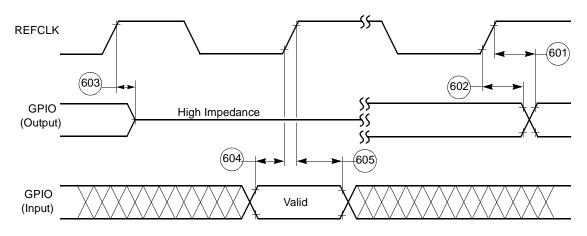


Figure 27. GPIO Timing

2.5.12 EE Signals

Table	29.	EE	Pin	Timing	

Number	Characteristics	Туре	Min	
65	EE0 (input)	Asynchronous	4 core clock periods	
66	EE1 (output)	Synchronous to Core clock	1 core clock period	

Notes: 1. The core clock is the SC140 core clock. The ratio between the core clock and CLKOUT is configured during power-on-reset.
 2. Refer to Table 1-4 on page 1-6 for details on EE pin functionality.

Figure 28 shows the signal behavior of the EE pins.

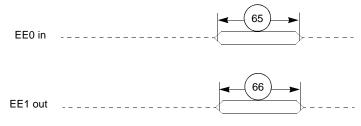


Figure 28. EE Pin Timing

2.5.13 JTAG Signals

Table 30. JTAG Timing

No.	Characteristics		All frequencies	
		Min	Max	
700	TCK frequency of operation $(1/(T_{c} \times 4); maximum 25 \text{ MHz})$	0.0	25	MHz
701	TCK cycle time	40.0	_	ns
702	TCK clock pulse width measured at V_{M} = 1.6 V			
	• High	20.0	—	ns
	• Low	16.0	—	ns
703	TCK rise and fall times	0.0	3.0	ns



Table 30. JTAG Timing (continued)

No.	Characteristics		All frequencies	
			Max	
704	Boundary scan input data set-up time	5.0		ns
705	Boundary scan input data hold time	20.0	_	ns
706	TCK low to output data valid	0.0	30.0	ns
707	TCK low to output high impedance	0.0	30.0	ns
708	TMS, TDI data set-up time	5.0	_	ns
709	TMS, TDI data hold time	20.0	_	ns
710	TCK low to TDO data valid	0.0	20.0	ns
711	TCK low to TDO high impedance	0.0	20.0	ns
712	TRST assert time	100.0	_	ns
713	TRST set-up time to TCK low	30.0	_	ns
Note:	All timings apply to OnCE module data transfers as well as any other transfers via the JTAG port.			

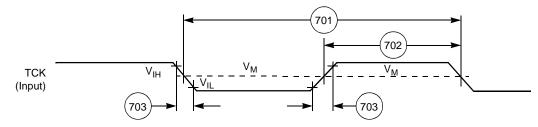


Figure 29. Test Clock Input Timing Diagram

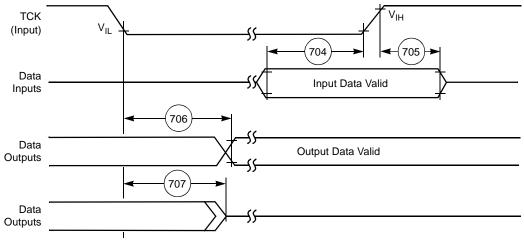


Figure 30. Boundary Scan (JTAG) Timing Diagram



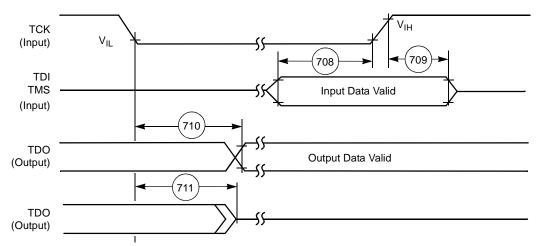
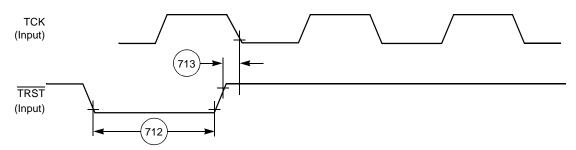


Figure 31. Test Access Port Timing Diagram





3 Hardware Design Considerations

The following sections discuss areas to consider when the MSC8122 device is designed into a system.

3.1 Start-up Sequencing Recommendations

Use the following guidelines for start-up and power-down sequences:

- Assert **PORESET** and **TRST** before applying power and keep the signals driven low until the power reaches the required minimum power levels. This can be implemented via weak pull-down resistors.
- CLKIN can be held low or allowed to toggle during the beginning of the power-up sequence. However, CLKIN must start toggling before the deassertion of PORESET and after both power supplies have reached nominal voltage levels.
- If possible, bring up V_{DD}/V_{CCSYN} and V_{DDH} together. If it is not possible, raise V_{DD}/V_{CCSYN} first and then bring up V_{DDH}. V_{DDH} should not exceed V_{DD}/V_{CCSYN} until V_{DD}/V_{CCSYN} reaches its nominal voltage level. Similarly, bring both voltage levels down together. If that is not possible reverse the power-up sequence, with V_{DDH} going down first and then V_{DD}/V_{CCSYN}.
- **Note:** This recommended power sequencing for the MSC8122 is different from the MSC8102. See Section 2.5.2 for start-up timing specifications.

External voltage applied to any input line must not exceed the I/O supply V_{DDH} by more than 0.8 V at any time, including during power-up. Some designs require pull-up voltages applied to selected input lines during power-up for configuration purposes. This is an acceptable exception to the rule. However, each such input can draw up to 80 mA per input pin per device in the system during start-up.

During the power-up sequence, if V_{DD} rises before V_{DDH} (see **Figure 6**), current can pass from the V_{DD} supply through the device ESD protection circuits to the V_{DDH} supply. The ESD protection diode can allow this to occur when V_{DD} exceeds V_{DDH} by more than 0.8 V. Design the power supply to prevent or minimize this effect using one of the following optional methods:

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- Never allow V_{DD} to exceed $V_{DDH} + 0.8V$.
- Design the V_{DDH} supply to prevent reverse current flow by adding a minimum 10 Ω resistor to GND to limit the current. Such a design yields an initial V_{DDH} level of $V_{DD} 0.8$ V before it is enabled.

After power-up, V_{DDH} must not exceed V_{DD}/V_{CCSYN} by more than 2.6 V.

3.2 Power Supply Design Considerations

When used as a drop-in replacement in MSC8102 applications or when implementing a new design, use the guidelines described in *Migrating Designs from the MSC8102 to the MSC8122* (AN2716) and the *MSC8122 Design Checklist* (AN3374 for optimal system performance. *MSC8122 and MSC8126 Power Circuit Design Recommendations and Examples* (AN2937) provides detailed design information. See Section 2.5.2 for start-up timing specifications.

Figure 33 shows the recommended power decoupling circuit for the core power supply. The voltage regulator and the decoupling capacitors should supply the required device current without any drop in voltage on the device pins. The voltage on the package pins should not drop below the minimum specified voltage level even for a very short spikes. This can be achieved by using the following guidelines:

- For the core supply, use a voltage regulator rated at 1.2 V with nominal rating of at least 3 A. This rating does not reflect actual average current draw, but is recommended because it resists changes imposed by transient spikes and has better voltage recovery time than supplies with lower current ratings.
- Decouple the supply using low-ESR capacitors mounted as close as possible to the socket. **Figure 33** shows three capacitors in parallel to reduce the resistance. Three capacitors is a recommended minimum number. If possible, mount at least one of the capacitors directly below the MSC8122 device.

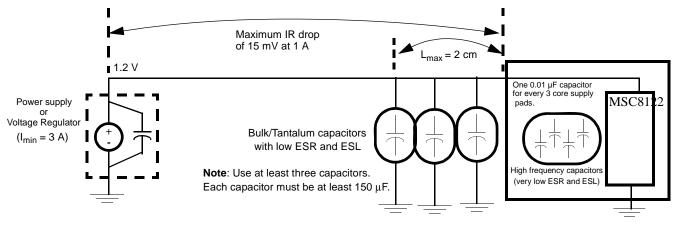


Figure 33. Core Power Supply Decoupling

Each V_{CC} and V_{DD} pin on the MSC8122 device should have a low-impedance path to the board power supply. Similarly, each GND pin should have a low-impedance path to the ground plane. The power supply pins drive distinct groups of logic on the chip. The V_{CC} power supply should have at least four 0.1 µF by-pass capacitors to ground located as closely as possible to the four sides of the package. The capacitor leads and associated printed circuit traces connecting to chip V_{CC} , V_{DD} , and GND should be kept to less than half an inch per capacitor lead. A four-layer board is recommended, employing two inner layers as V_{CC} and GND planes.

All output pins on the MSC8122 have fast rise and fall times. PCB trace interconnection length should be minimized to minimize undershoot and reflections caused by these fast output switching times. This recommendation particularly applies to the address and data buses. Maximum PCB trace lengths of six inches are recommended. For the DSI control signals in synchronous mode, ensure that the layout supports the DSI AC timing requirements and minimizes any signal crosstalk. Capacitance calculations should consider all device loads as well as parasitic capacitances due to the PCB traces. Attention to proper PCB layout and bypassing becomes especially critical in systems with higher capacitive loads because these loads create higher transient currents in the V_{CC}, V_{DD}, and GND circuits. Pull up all unused inputs or signals that will be inputs during reset.

Special care should be taken to minimize the noise levels on the PLL supply pins. There is one pair of PLL supply pins: V_{CCSYN} -GND_{SYN}. To ensure internal clock stability, filter the power to the V_{CCSYN} input with a circuit similar to the one in



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Figure 34. For optimal noise filtering, place the circuit as close as possible to V_{CCSYN} . The 0.01- μ F capacitor should be closest to V_{CCSYN} , followed by the 10- μ F capacitor, the 10-nH inductor, and finally the 10- Ω resistor to V_{DD} . These traces should be kept short and direct. Provide an extremely low impedance path to the ground plane for GND_{SYN} . Bypass GND_{SYN} to V_{CCSYN} by a 0.01- μ F capacitor located as close as possible to the chip package. For best results, place this capacitor on the backside of the PCB aligned with the depopulated void on the MSC8122 located in the square defined by positions, L11, L12, L13, M11, M12, M13, N11, N12, and N13.

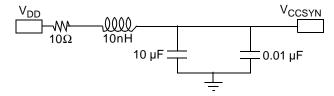


Figure 34. V_{CCSYN} Bypass

3.3 Connectivity Guidelines

Unused output pins can be disconnected, and unused input pins should be connected to the non-active value, via resistors to V_{DDH} or GND, except for the following:

- If the DSI is unused (DDR[DSIDIS] is set), HCS and HBCS must pulled up and all the rest of the DSI signals can be disconnected.
- When the DSI uses synchronous mode, HTA must be pulled up. In asynchronous mode, HTA should be pulled either up or down, depending on design requirements.
- HDST can be disconnected if the DSI is in big-endian mode, or if the DSI is in little-endian mode and the DCR[DSRFA] bit is set.
- When the DSI is in 64-bit data bus mode and DCR[BEM] is cleared, pull up HWBS[1-3]/HDBS[1-3]/HWBE[1-3]/HDBE[1-3]/HDBE[1-3]/HDBE[4-7]/HDBE[4-7]/PWE[4-7]/PWE[4-7]/PSDDQM[4-7]/PBS[4-7].
- When the DSI is in 32-bit data bus mode and DCR[BEM] is cleared, HWBS[1-3]/HDBS[1-3]/HDBE[1-3]
- When the DSI is in asynchronous mode, HBRST and HCLKIN should either be disconnected or pulled up.
- When the DSI uses sliding window address mode (DCR[SLDWA] = 1), the external HA[11–13] signals must be connected (tied) to the correct voltage levels so that the host can perform the first access to the DCR. After reset, the DSI expects full address mode (DCR[SLDWA] = 0). The DCR address in the DSI memory map is 0x1BE000, which requires the following connections:
 - HA11 must be pulled high (1)
 - HA12 must be pulled high (1)
 - HA13 must be pulled low (0)
- The following signals must be pulled up: HRESET, SRESET, ARTRY, TA, TEA, PSDVAL, and AACK.
- In single-master mode (BCR[EBM] = 0) with internal arbitration (PPC_ACR[EARB] = 0):
 - BG, DBG, and TS can be left unconnected.
 - EXT_BG[2–3], EXT_DBG[2–3], and GBL can be left unconnected if they are multiplexed to the system bus functionality. For any other functionality, connect the signal lines based on the multiplexed functionality.
 - BR must be pulled up.
 - EXT_BR[2–3] must be pulled up if multiplexed to the system bus functionality.
- If there is an external bus master (BCR[EBM] = 1):
 - BR, BG, DBG, and TS must be pulled up.
 - EXT_BR[2-3], EXT_BG[2-3], and EXT_DBG[2-3] must be pulled up if multiplexed to the system bus functionality.
- In single-master mode, ABB and DBB can be selected as IRQ inputs and be connected to the non-active value. In other modes, they must be pulled up.

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- **Note:** The MSC8122 does not support DLL-enabled mode. For the following two clock schemes, ensure that the DLL is disabled (that is, the DLLDIS bit in the Hard Reset Configuration Word is set).
 - If no system synchronization is required (for example, the design does not use SDRAM), you can use any of the available clock modes.
 - In the CLKIN synchronization mode, use the following connections:
 - Connect the oscillator output through a buffer to CLKIN.
 - Connect the CLKIN buffer output to the slave device (for example, SDRAM) making sure that the delay path between the clock buffer to the MSC8122 and the SDRAM is equal (that is, has a skew less than 100 ps).
 Valid clock modes in this scheme area 0.7, 15, 10, 21, 22, 28, 20, 20, and 21.
 - Valid clock modes in this scheme are: 0, 7, 15, 19, 21, 23, 28, 29, 30, and 31.
 - In CLKOUT synchronization mode (for 1.2 V devices), CLKOUT is the main clock to SDRAM. Use the following connections:
 - Connect the oscillator output through a buffer to CLKIN.
 - Connect CLKOUT through a zero-delay buffer to the slave device (for example, SDRAM) using the following guidelines:
 - The maximum delay between the slave and CLKOUT must not exceed 0.7 ns.
 - The maximum load on CLKOUT must not exceed 10 pF.
 - Use a zero-delay buffer with a jitter less than 0.3 ns.
 - All clock modes are valid in this clock scheme.

Note: See the Clock chapter in the *MSC8122 Reference Manual* for details.

- If the 60x-compatible system bus is not used and SIUMCR[PBSE] is set, PPBS can be disconnected. Otherwise, it should be pulled up.
- The following signals: SWTE, DSISYNC, DSI64, MODCK[1–2], CNFGS, CHIPID[0–3], RSTCONF and BM[0–2] are
 used to configure the MSC8122 and are sampled on the deassertion of the PORESET signal. Therefore, they should
 be tied to GND or V_{DDH} or through a pull-down or a pull-up resistor until the deassertion of the PORESET signal.
- When they are used, INT_OUT (if SIUMCR[INTODC] is cleared), NMI_OUT, and IRQxx (if not full drive) signals must be pulled up.
- When the Ethernet controller is enabled and the SMII mode is selected, GPIO10 and GPIO14 must not be connected externally to any signal line.
- **Note:** For details on configuration, see the *MSC8122 User's Guide* and *MSC8122 Reference Manual*. For additional information, refer to the *MSC8122 Design Checklist* (AN2787).

3.4 External SDRAM Selection

The external bus speed implemented in a system determines the speed of the SDRAM used on that bus. However, because of differences in timing characteristics among various SDRAM manufacturers, you may have use a faster speed rated SDRAM to assure efficient data transfer across the bus. For example, for 166 MHz operation, you may have to use 183 or 200 MHz SDRAM. Always perform a detailed timing analysis using the MSC8122 bus timing values and the manufacturer specifications for the SDRAM to ensure correct operation within your system design. The output delay listed in SDRAM specifications is usually given for a load of 30 pF. Scale the number to your specific board load using the typical scaling number provided by the SDRAM manufacturer.

3.5 Thermal Considerations

An estimation of the chip-junction temperature, T_J, in °C can be obtained from the following:

$$T_J = T_A + (R_{\bigcup JA} \times P_D)$$
 Eqn. 1

where

$$\begin{split} T_A &= \text{ambient temperature near the package (°C)} \\ R_{\Theta JA} &= \text{junction-to-ambient thermal resistance (°C/W)} \\ P_D &= P_{INT} + P_{I/O} = \text{power dissipation in the package (W)} \\ P_{INT} &= I_{DD} \times V_{DD} = \text{internal power dissipation (W)} \\ P_{I/O} &= \text{power dissipated from device on output pins (W)} \end{split}$$

The power dissipation values for the MSC8122 are listed in **Table 2-3**. The ambient temperature for the device is the air temperature in the immediate vicinity that would cool the device. The junction-to-ambient thermal resistances are JEDEC standard values that provide a quick and easy estimation of thermal performance. There are two values in common usage: the value determined on a single layer board and the value obtained on a board with two planes. The value that more closely approximates a specific application depends on the power dissipated by other components on the printed circuit board (PCB). The value obtained using a single layer board is appropriate for tightly packed PCB configurations. The value obtained using a board with internal planes is more appropriate for boards with low power dissipation (less than 0.02 W/cm² with natural convection) and well separated components. Based on an estimation of junction temperature using this technique, determine whether a more detailed thermal analysis is required. Standard thermal management techniques can be used to maintain the device thermal junction temperature below its maximum. If T_J appears to be too high, either lower the ambient temperature or the power dissipation of the chip. You can verify the junction temperature by measuring the case temperature using a small diameter thermocouple (40 gauge is recommended) or an infrared temperature sensor on a spot on the device case that is painted black. The MSC8122 device case surface is too shiny (low emissivity) to yield an accurate infrared temperature measurement. Use the following equation to determine T_J:

$$T_J = T_T + (\theta_{JA} \times P_D)$$
 Eqn. 2

where

 T_T = thermocouple (or infrared) temperature on top of the package (°C)

 θ_{JA} = thermal characterization parameter (°C/W)

 P_D = power dissipation in the package (W)

Note: See MSC8102, MSC8122, and MSC8126 Thermal Management Design Guidelines (AN2601/D).

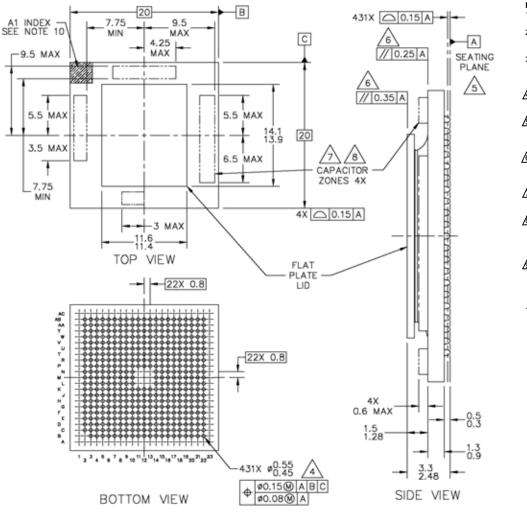
4 Ordering Information

Consult a Freescale Semiconductor sales office or authorized distributor to determine product availability and place an order.

		Core	Operating Temperature	Core Frequency (MHz)	Order Number		
Part	Package Type	Voltage			Lead-Free	Lead-Bearing	
MSC8122	Flip Chip Plastic Ball Grid Array (FC-PBGA)	1.1 V	-40° to 105°C	300	MSC8122TVT4800V	MSC8122TMP4800V	
				400	MSC8122TVT6400V	MSC8122TMP6400V	
		1.2 V	–40° to 105°C	400	MSC8122TVT6400	MSC8122TMP6400	
			0° to 90°C	500	MSC8122VT8000	MSC8122MP8000	



Package Information 5



- Notes: 1. All dimensions in millimeters.
- 2. Dimensioning and tolerancing per ASME Y14.5M–1994.
- 3. Features are symmetrical about the package center lines unless dimensioned otherwise.
- A Maximum solder ball diameter measured parallel to Datum A.
- Datum A, the seating plane, is determined by the spherical crowns of the solder balls.

A Parallelism measurement shall exclude any effect of mark on top surface of package.

Capacitors may not be present on all devices.

Caution must be taken not to short capacitors or exposed metal capacitor pads on package top.

FC CBGA (Ceramic) package code: 5238. FC PBGA (Plastic) package code: 5263.

10.Pin 1 indicator can be in the form of number 1 marking or an "L" shape marking.



6 **Product Documentation**

- MSC8122 Technical Data Sheet (MSC8122). Details the signals, AC/DC characteristics, clock signal characteristics, package and pinout, and electrical design considerations of the MSC8122 device.
- MSC8122 Reference Manual (MSC8122RM). Includes functional descriptions of the extended cores and all the internal subsystems including configuration and programming information.
- Application Notes. Cover various programming topics related to the StarCore DSP core and the MSC8122 device.
- SC140 DSP Core Reference Manual. Covers the SC140 core architecture, control registers, clock registers, program control, and instruction set.

7 Revision History

Table 31 provides a revision history for this data sheet.

Table 31. Document Revision History

Revision	Date	Description
0	May 2004	Initial release.
1	Jun. 2004	Updated timing number 32b.Updated DSI timing specifications.
2	Sep 2004	 New orderable parts added with other core voltage and temperature options. Updated thermal characteristics. In Table 2-14, removed references to 30 pF. Design guidelines and layout recommendations updated.
3	Nov. 2004	 Added 500 MHz core and 166 MHz bus speed options. Definitions of GPIO[27–28] updated. Bus, TDM, and GPIO timing updated. I²C timing changed to GPIO timing. GPIO[27–28] connections updated. MWBEn replaced with correct name HWBEn. Design guidelines update.
4	Jan. 2005	 Package type changed to FC-PBGA for all frequencies. Low-voltage 300 MHz power changed to 1.1 V. HRESET and SRESET definitions updated. Undershoot and overshoot values added for V_{DDH}. RMII timing updated. Design guidelines updated and reorganized.
5	Apr. 2005	 Added 400 MHz, 1.1 V core part. Temperature range descriptions changed to standard and extended. CLKOUT timing specifications added. Device start-up guidelines added to design considerations and updated power supply guidelines. Ordering information updated.
6	May 2005	Multiple AC timing specifications updated.
7	May 2005	Multiple AC timing specifications updated.
8	Jul. 2005	Multiple AC timing specifications updated.
9	Jul. 2005	AC specification table layout modified.
10	Sep. 2005	 ETHTX_EN type and TRST description updated. Package drawing updated. Clock specifications updated. Start-up sequence updated.
11	Oct 2005	 V_{DDH} + 10% changed to V_{DDH} + 8% in Figure 2-1. V_{DDH} +20% changed to V_{DDH} + 17% in Figure 2-1.
12	Apr 2006	• Reset timing updated to reflect actual values in Table 2-11.
13	Oct. 2006	Added new timings 17 and 18 for IRQ set time and pulse width in Table 2-13
14	Dec. 2007	 Converted to new data sheet format. Added PLL supply current to Table 5 in Section 2.4. Modified Figure 5 in Section 2.4 to make it clear that the time limits for undershoot referred to values below -0.3 V and not GND. Added cross-references between Sections 2.5.2 and Section 3.1 and 3.2. Added power-sequence guidelines to Sections 2.5.2. Added CLKIN jitter characteristic specifications to Table 9. Added additional guidelines to prevent reverse current to Section 3.1. Added connectivity guidelines for DSI in sliding windows mode to Section 3.3.
15	May 2008	• Changed V _{IL} maximum and reference value to 0.8 V in Table 5 .
16	Dec. 2008	• Clarified the wording of note 2 in Table 15 on p. 24.



MSC8122 Quad Digital Signal Processor Data Sheet, Rev. 16



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