

PIC18F85J90 Family Data Sheet

64/80-Pin, High-Performance Microcontrollers with LCD Driver and nanoWatt Technology

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64/80-Pin, High-Performance Microcontrollers with LCD Driver and nanoWatt Technology

LCD Driver Module Features:

- Direct LCD Panel Drive Capability:
- Can drive LCD panel while in Sleep mode
 Up to 48 Segments and 192 Pixels; Software
- Selectable
- Programmable LCD Timing module:
 - Multiple LCD timing sources available
 - Up to 4 commons: static, 1/2, 1/3 or 1/4 multiplex
 - Static, 1/2 or 1/3 Bias configuration
- Integrated Charge-Pump Module with Voltage Boost

Special Microcontroller Features:

- 1000 Erase/Write Cycle Flash Program Memory, typical
- Flash Retention: 20 Years Minimum
- Self-Programmable under Software Control
- Priority Levels for Interrupts
- 8 x 8 Single-Cycle Hardware Multiplier
- Extended Watchdog Timer (WDT):
- Programmable period from 4 ms to 131s
- In-Circuit Serial Programming[™] (ICSP[™]) via two pins
- In-Circuit Debug with 3 Breakpoints via two pins
- Operating Voltage Range: 2.0V to 3.6V
- On-Chip 2.5V Regulator

Flexible Oscillator Structure:

- Two Crystal modes, 4-25 MHz
- Two External Clock modes, up to 40 MHz
- Internal Oscillator Block:
 - 8 user-selectable frequencies from 31.25 kHz to 8 MHz
- Secondary Oscillator using Timer1 @ 32 kHz
- Fail-Safe Clock Monitor:
 - Allows for safe shutdown if peripheral clock fails

Low-Power Features:

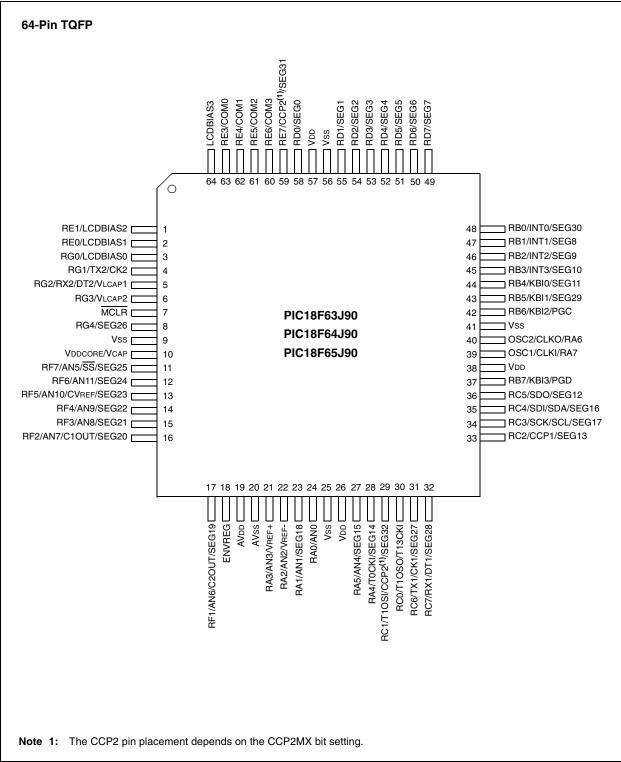
- Power-Managed modes: Run, Idle, Sleep
- Run current down to 9 μA, typical
- Idle current down to 2.5 µA, typical
- Sleep current down to 0.1 µA, typical
- Fast INTOSC startup from SLEEP
- Two-Speed Oscillator Start-up reduces crystal stabilization wait time

Peripheral Highlights:

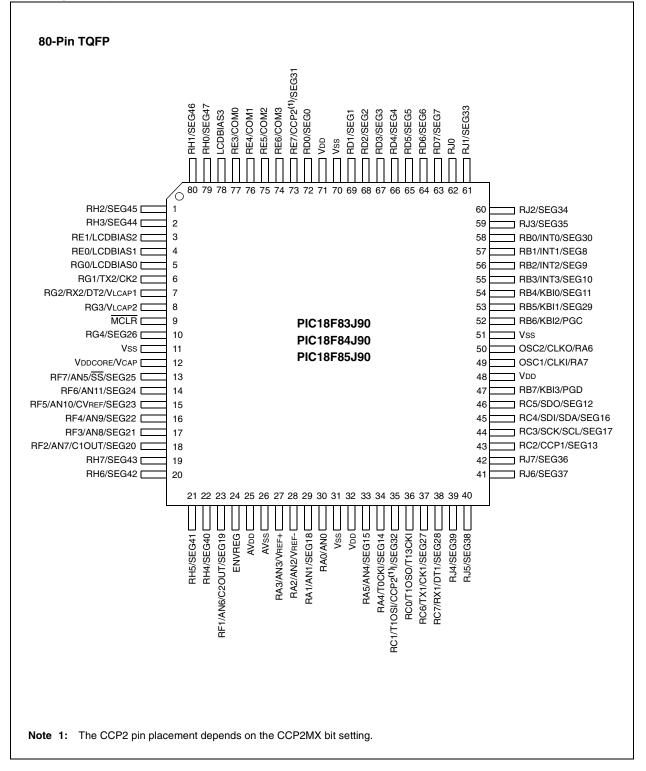
- High-Current Sink/Source: 25 mA/25 mA (PORTB and PORTC)
- Sleep current as low as 100nA
- Up to Four External Interrupts
- Four 8-Bit/16-Bit Timer/Counter modules
 Uses Timer1
- Two Capture/Compare/PWM (CCP) modules:
 - Capture is 16-bit, max. resolution 6.25 ns (TCY/16)
 - Compare is 16-bit, max. resolution 100 ns (TCY)
 - PWM output: PWM resolution is up to 10-bit
- Master Synchronous Serial Port (MSSP) module with two modes of Operation:
 - 3-wire/4-wire SPI (supports all 4 SPI modes)
 - I²C[™] Master and Slave mode
- One Addressable USART module
- One Enhanced USART module:
 - Supports LIN 1.2
 - Auto-wake-up on Start bit and Break character
 - Auto-Baud Detect
- 10-Bit, up to 12-Channel A/D Converter:
 - Auto-acquisition
 - Conversion available during Sleep
- Two Analog Comparators
- Programmable Reference Voltage for Comparators

	Prog	ram Memory	SRAM			ł		M	SSP	т/ Т		ors	
Device	Flash (bytes)	# Single-Word Instructions	Data Memory (bytes)	I/O	LCD (Pixels)	Timers 8/16-Bit	ССР	SPI	Master I ² C™	EUSAR'	10-Bit A/D (ch)	Comparators	BOR/ LVD
PIC18F63J90	8K	4096	1024	51	132	1/3	2	Y	Y	1/1	12	2	Y
PIC18F64J90	16K	8192	1024	51	132	1/3	2	Y	Y	1/1	12	2	Y
PIC18F65J90	32K	16384	2048	51	132	1/3	2	Y	Y	1/1	12	2	Y
PIC18F83J90	8K	4096	1024	67	192	1/3	2	Y	Y	1/1	12	2	Y
PIC18F84J90	16K	8192	1024	67	192	1/3	2	Y	Y	1/1	12	2	Y
PIC18F85J90	32K	16384	2048	67	192	1/3	2	Y	Y	1/1	12	2	Y

Pin Diagrams



Pin Diagrams (Continued)



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NOTES:

1.0 DEVICE OVERVIEW

This document contains device specific information for the following devices:

- PIC18F63J90 PIC18F83J90
- PIC18F64J90
- PIC18F84J90
- PIC18F65J90
- PIC18F85J90

This family combines the traditional advantages of all PIC18 microcontrollers – namely, high computational performance and a rich feature set – with a versatile on-chip LCD driver, while maintaining an extremely competitive price point. These features make the PIC18F85J90 family a logical choice for many high-performance applications where price is a primary consideration.

1.1 Core Features

1.1.1 nanoWatt TECHNOLOGY

All of the devices in the PIC18F85J90 family incorporate a range of features that can significantly reduce power consumption during operation. Key items include:

- Alternate Run Modes: By clocking the controller from the Timer1 source or the internal RC oscillator, power consumption during code execution can be reduced by as much as 90%.
- Multiple Idle Modes: The controller can also run with its CPU core disabled but the peripherals still active. In these states, power consumption can be reduced even further, to as little as 4% of normal operation requirements.
- **On-the-Fly Mode Switching:** The power-managed modes are invoked by user code during operation, allowing the user to incorporate power-saving ideas into their application's software design.

1.1.2 OSCILLATOR OPTIONS AND FEATURES

All of the devices in the PIC18F85J90 family offer six different oscillator options, allowing users a range of choices in developing application hardware. These include:

- Two Crystal modes, using crystals or ceramic resonators.
- Two External Clock modes, offering the option of a divide-by-4 clock output.
- A Phase Lock Loop (PLL) frequency multiplier, available to the External Oscillator modes which allows clock speeds of up to 40 MHz.
- An internal oscillator block which provides an 8 MHz clock (±2% accuracy) and an INTRC source (approximately 31 kHz, stable over temperature and VDD), as well as a range of six user-selectable clock frequencies, between 125 kHz to 4 MHz, for a total of eight clock frequencies. This option frees the two oscillator pins for use as additional general purpose I/O.

The internal oscillator block provides a stable reference source that gives the family additional features for robust operation:

- Fail-Safe Clock Monitor: This option constantly monitors the main clock source against a reference signal provided by the internal oscillator. If a clock failure occurs, the controller is switched to the internal oscillator, allowing for continued low-speed operation or a safe application shutdown.
- **Two-Speed Start-up:** This option allows the internal oscillator to serve as the clock source from Power-on Reset, or wake-up from Sleep mode, until the primary clock source is available.

1.1.3 MEMORY OPTIONS

The PIC18F85J90 family provides a range of program memory options, from 8 Kbytes to 32 Kbytes of code space. The Flash cells for program memory are rated to last up to 1000 erase/write cycles. Data retention without refresh is conservatively estimated to be greater than 20 years.

The PIC18F85J90 family also provides plenty of room for dynamic application data, with up to 2048 bytes of data RAM.

1.1.4 EXTENDED INSTRUCTION SET

The PIC18F85J90 family implements the optional extension to the PIC18 instruction set, adding 8 new instructions and an Indexed Addressing mode. Enabled as a device configuration option, the extension has been specifically designed to optimize re-entrant application code originally developed in high-level languages, such as 'C'.

1.1.5 EASY MIGRATION

Regardless of the memory size, all devices share the same rich set of peripherals, allowing for a smooth migration path as applications grow and evolve.

The consistent pinout scheme used throughout the entire family also aids in migrating to the next larger device. This is true when moving between the 64-pin members, between the 80-pin members, or even jumping from 64-pin to 80-pin devices.

The PIC18F85J90 family is also largely pin compatible with other PIC18 families, such as the PIC18F8720 and PIC18F8722, as well as the PIC18F8490 family of microcontrollers with LCD drivers. This allows a new dimension to the evolution of applications, allowing developers to select different price points within Microchip's PIC18 portfolio, while maintaining a similar feature set.

1.2 LCD Driver

The on-chip LCD driver includes many features that make the integration of displays in low-power applications easier. These include an integrated voltage regulator with charge pump that allows contrast control in software and display operation above device VDD.

1.3 Other Special Features

- **Communications:** The PIC18F85J90 family incorporates a range of serial communication peripherals, including an Addressable USART, a separate Enhanced USART that supports LIN specification 1.2, and one Master SSP module capable of both SPI and I²CTM (Master and Slave) modes of operation.
- CCP Modules: All devices in the family incorporate two Capture/Compare/PWM (CCP) modules. Up to four different time bases may be used to perform several different operations at once.
- **10-Bit A/D Converter:** This module incorporates programmable acquisition time, allowing for a channel to be selected and a conversion to be initiated without waiting for a sampling period and thus, reducing code overhead.
- Extended Watchdog Timer (WDT): This enhanced version incorporates a 16-bit prescaler, allowing an extended time-out range that is stable across operating voltage and temperature. See Section 25.0 "Electrical Characteristics" for time-out periods.

1.4 Details on Individual Family Members

Devices in the PIC18F85J90 family are available in 64-pin and 80-pin packages. Block diagrams for the two groups are shown in Figure 1-1 and Figure 1-2.

The devices are differentiated from each other in four ways:

- 1. Flash program memory (three sizes, ranging from 8 Kbytes for PIC18FX3J90 devices to 32 Kbytes for PIC18FX5J90 devices).
- Data RAM (1024 bytes for PIC18FX3J90 and PIC18FX4J90 devices, 2048 bytes for PIC18FX5J90 devices).
- I/O ports (7 bidirectional ports on 64-pin devices, 9 bidirectional ports on 80-pin devices).
- LCD Pixels: 132 pixels (33 SEGs x 4 COMs) can be driven by 64-pin devices; 192 pixels (48 SEGs x 4 COMs) can be driven by 80-pin devices.

All other features for devices in this family are identical. These are summarized in Table 1-1 and Table 1-2.

The pinouts for all devices are listed in Table 1-3 and Table 1-4.

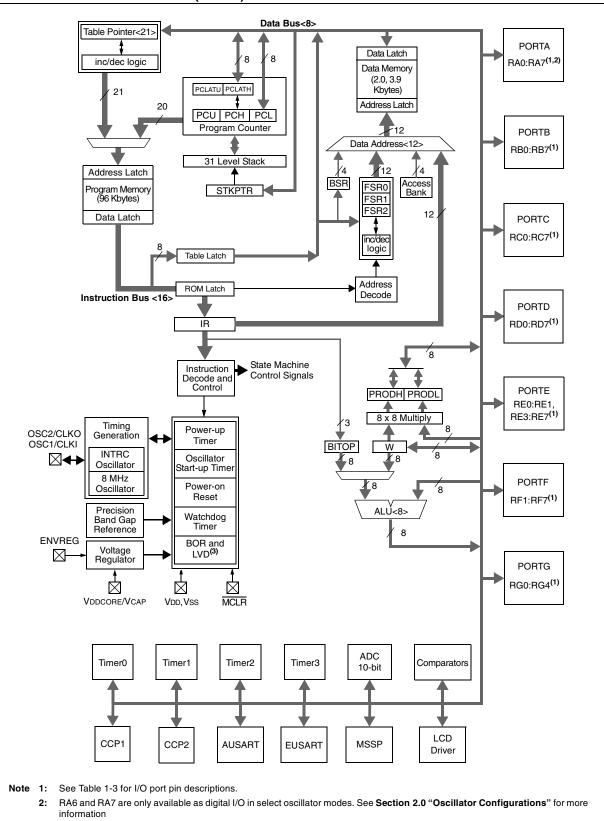
Features	PIC18F63J90	PIC18F64J90	PIC18F65J90				
Operating Frequency		DC – 40 MHz					
Program Memory (Bytes)	8K	16K	32K				
Program Memory (Instructions)	4096	8192	16384				
Data Memory (Bytes)	1024	1024	2048				
Interrupt Sources		27					
I/O Ports		Ports A, B, C, D, E, F, G					
LCD Driver (available pixels to drive)	132 (33 SEGs x 4 COMs)						
Timers	4						
Capture/Compare/PWM Modules		2					
Serial Communications	MSSP, Ad	dressable USART, Enhand	ed USART				
10-bit Analog-to-Digital Module		12 Input Channels					
Resets (and Delays)	POR, BOR, RESET Instruction, Stack Full, Stack Underflow, MCLR, WDT (PWRT, OST)						
Instruction Set	75 Instructions, 83 with Extended Instruction Set enabled						
Packages		64-pin TQFP					

TABLE 1-1: DEVICE FEATURES FOR THE PIC18F85J90 FAMILY (64-PIN DEVICES)

TABLE 1-2: DEVICE FEATURES FOR THE PIC18F85J90 FAMILY (80-PIN DEVICES)

Features	PIC18F83J90	PIC18F84J90	PIC18F85J90	
Operating Frequency		DC – 40 MHz		
Program Memory (Bytes)	8K	16K	32K	
Program Memory (Instructions)	4096	8192	16384	
Data Memory (Bytes)	1024	1024	2048	
Interrupt Sources		27		
I/O Ports	F	Ports A, B, C, D, E, F, G, H,	J	
LCD Driver (available pixels to drive)		192 (48 SEGs x 4 COMs))	
Timers		4		
Capture/Compare/PWM Modules		2		
Serial Communications	MSSP, Ad	ldressable USART, Enhand	ed USART	
10-bit Analog-to-Digital Module		12 Input Channels		
Resets (and Delays)	POR, BOR, RESET Instruction, Stack Full, Stack Underflow, MCLR, WD (PWRT, OST)			
Instruction Set	75 Instructions	, 83 with Extended Instruct	ion Set enabled	
Packages		80-pin TQFP		





3: Brown-out Reset and Low-Voltage Detect functions are provided when the on-board voltage regulator is enabled.

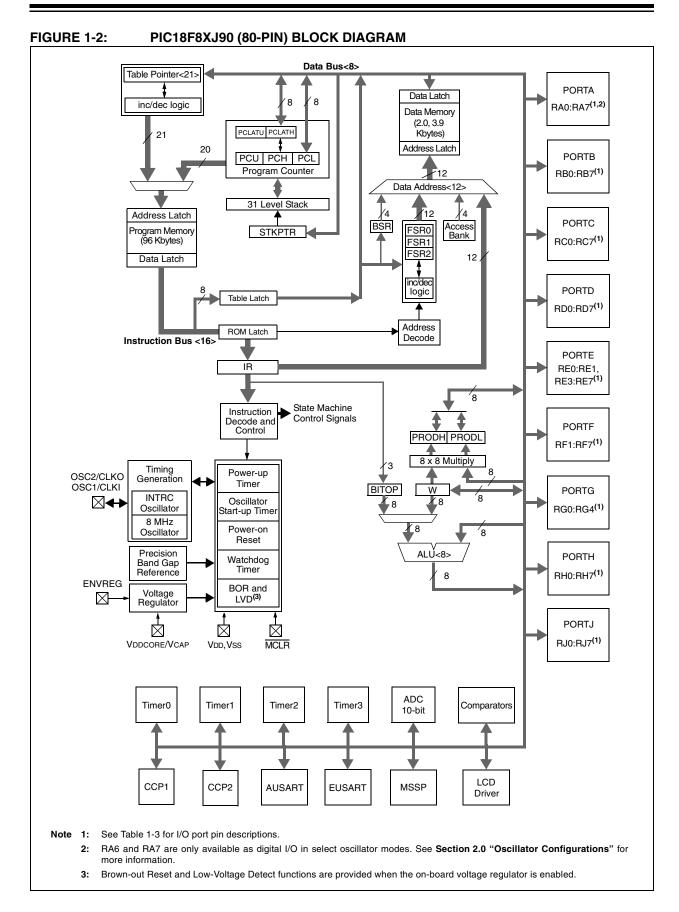


TABLE 1-3:PIC18F6XJ90 PINOUT I/O DESCRIPTIONS

Pin Name	Pin Number	Pin	Buffer	Description
	TQFP	Туре	Туре	· · · · · · · · · · · · · · · · · · ·
MCLR	7	I	ST	Master Clear (input) or programming voltage (input). This pin is an active-low Reset to the device.
OSC1/CLKI/RA7 OSC1 CLKI RA7	39	 /O	CMOS CMOS TTL	Oscillator crystal or external clock input. Oscillator crystal input. External clock source input. Always associated with pin function OSC1. (See related OSC1/CLKI, OSC2/CLKO pins.) General purpose I/O pin.
OSC2/CLKO/RA6 OSC2 CLKO	40	0	_	Oscillator crystal or clock output. Oscillator crystal output. Connects to crystal or resonator in Crystal Oscillator mode. In EC modes, OSC2 pin outputs CLKO, which has
RA6		I/O	TTL	1/4 the frequency of OSC1 and denotes the instruction cycle rate. General purpose I/O pin.
RA0/AN0 RA0 AN0	24	I/O I	TTL Analog	PORTA is a bidirectional I/O port. Digital I/O. Analog input 0.
RA1/AN1/SEG18 RA1 AN1 SEG18	23	I/O I O	TTL Analog Analog	Digital I/O. Analog input 1. SEG18 output for LCD.
RA2/AN2/VREF- RA2 AN2 VREF-	22	I/O I I	TTL Analog Analog	Digital I/O. Analog input 2. A/D reference voltage (Low) input.
RA3/AN3/VREF+ RA3 AN3 VREF+	21	I/O I I	TTL Analog Analog	Digital I/O. Analog input 3. A/D reference voltage (High) input.
RA4/T0CKI/SEG14 RA4 T0CKI SEG14	28	I/O I O	ST/OD ST Analog	Digital I/O. Open-drain when configured as output. Timer0 external clock input. SEG14 output for LCD.
RA5/AN4/SEG15 RA5 AN4 SEG15	27	I/O I O	TTL Analog Analog	Digital I/O. Analog input 4. SEG15 output for LCD.
RA6				See the OSC2/CLKO/RA6 pin.
RA7				See the OSC1/CLKI/RA7 pin.
ST = Schn I = Input P = Powe	er			CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD) Configuration bit is set.

Note 1: Default assignment for CCP2 when CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin	Buffer Type	Description	
i in Name	TQFP	Туре		Description	
				PORTB is a bidirectional I/O port. PORTB can be software programmed for internal weak pull-ups on all inputs.	
RB0/INT0/SEG30 RB0 INT0 SEG30	48	I/O I O	TTL ST Analog	Digital I/O. External interrupt 0. SEG30 output for LCD.	
RB1/INT1/SEG8 RB1 INT1 SEG8	47	I/O I O	TTL ST Analog	Digital I/O. External interrupt 1. SEG8 output for LCD.	
RB2/INT2/SEG9 RB2 INT2 SEG9	46	I/O I O	TTL ST Analog	Digital I/O. External interrupt 2. SEG9 output for LCD.	
RB3/INT3/SEG10 RB3 INT3 SEG10	45	I/O I O	TTL ST Analog	Digital I/O. External interrupt 3. SEG10 output for LCD.	
RB4/KBI0/SEG11 RB4 KBI0 SEG11	44	I/O I O	TTL TTL Analog	Digital I/O. Interrupt-on-change pin. SEG11 output for LCD.	
RB5/KBI1/SEG29 RB5 KBI1 SEG29	43	I/O I O	TTL TTL Analog	Digital I/O. Interrupt-on-change pin. SEG29 output for LCD.	
RB6/KBI2/PGC RB6 KBI2 PGC	42	I/O I I/O	TTL TTL ST	Digital I/O. Interrupt-on-change pin. In-Circuit Debugger and ICSP™ programming clock pin	
RB7/KBI3/PGD RB7 KBI3 PGD	37	I/O I I/O	TTL TTL ST	Digital I/O. Interrupt-on-change pin. In-Circuit Debugger and ICSP programming data pin.	
I = Inpu P = Pow	nitt Trigger input t er	with C		CMOS = CMOS compatible input or output els Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD) Configuration bit is set.	

TABLE 1-3:	PIC18F6XJ90 PINOUT I/O DESCRIPTIONS ((CONTINUED)

Note 1: Default assignment for CCP2 when CCP2MX Configuration bit is set.

Din Nama	Pin Number	Pin	Buffer	Deceriation		
Pin Name	TQFP	Туре	Туре	Description		
				PORTC is a bidirectional I/O port.		
RC0/T1OSO/T13CKI RC0 T1OSO T13CKI	30	I/O O I	ST — ST	Digital I/O. Timer1 oscillator output. Timer1/Timer3 external clock input.		
RC1/T1OSI/CCP2/SEG32 RC1 T1OSI CCP2 ⁽¹⁾ SEG32	29	I/O I I/O O	ST CMOS ST Analog	Digital I/O. Timer1 oscillator input. Capture2 input/Compare2 output/PWM2 output. SEG32 output for LCD.		
RC2/CCP1/SEG13 RC2 CCP1 SEG13	33	I/O I/O O	ST ST Analog	Digital I/O. Capture1 input/Compare1 output/PWM1 output. SEG13 output for LCD.		
RC3/SCK/SCL/SEG17 RC3 SCK SCL SEG17	34	I/O I/O I/O O	ST ST ST Analog	Digital I/O. Synchronous serial clock input/output for SPI mode. Synchronous serial clock input/output for I ² C™ mode. SEG17 output for LCD.		
RC4/SDI/SDA/SEG16 RC4 SDI SDA SEG16	35	I/O I I/O O	ST ST ST Analog	Digital I/O. SPI data in. I ² C data I/O. SEG16 output for LCD.		
RC5/SDO/SEG12 RC5 SDO SEG12	36	I/O O O	ST Analog	Digital I/O. SPI data out. SEG12 output for LCD.		
RC6/TX1/CK1/SEG27 RC6 TX1 CK1 SEG27	31	I/O O I/O O	ST — ST Analog	Digital I/O. EUSART asynchronous transmit. EUSART synchronous clock (see related RX1/DT1). SEG27 output for LCD.		
RC7/RX1/DT1/SEG28 RC7 RX1 DT1 SEG28	32	I/O I /O O	ST ST ST Analog	Digital I/O. EUSART asynchronous receive. EUSART synchronous data (see related TX1/CK1). SEG28 output for LCD.		
Legend: TTL = TTL compatible input CMOS = CMOS compatible input or output ST = Schmitt Trigger input with CMOS levels Analog = Analog input I = Input O = Output P = Power OD = Open-Drain (no P diode to VDD) Note 1: Default assignment for CCP2 when CCP2MX Configuration bit is set.						

TABLE 1-3: PIC18F6XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)

Din Nome	Pin Number	Pin	Buffer	Description
Pin Name	TQFP	Туре	Туре	Description
				PORTD is a bidirectional I/O port.
RD0/SEG0 RD0 SEG0	58	I/O O	ST Analog	Digital I/O. SEG0 output for LCD.
RD1/SEG1 RD1 SEG1	55	I/O O	ST Analog	Digital I/O. SEG1 output for LCD.
RD2/SEG2 RD2 SEG2	54	I/O O	ST Analog	Digital I/O. SEG2 output for LCD.
RD3/SEG3 RD3 SEG3	53	I/O O	ST Analog	Digital I/O. SEG3 output for LCD.
RD4/SEG4 RD4 SEG4	52	I/O O	ST Analog	Digital I/O. SEG4 output for LCD.
RD5/SEG5 RD5 SEG5	51	I/O O	ST Analog	Digital I/O. SEG5 output for LCD.
RD6/SEG6 RD6 SEG6	50	I/O O	ST Analog	Digital I/O. SEG6 output for LCD.
RD7/SEG7 RD7 SEG7	49	I/O O	ST Analog	Digital I/O. SEG7 output for LCD.
	compatible input hitt Trigger input	with C	MOS leve	CMOS = CMOS compatible input or output Analog = Analog input O = Output

TABLE 1-3: PIC18F6XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)

Ρ = Power

Note 1: Default assignment for CCP2 when CCP2MX Configuration bit is set.

Din Nome	Pin Number	Pin	Buffer	Description
Pin Name	TQFP	Туре	Туре	Description
				PORTE is a bidirectional I/O port.
RE0/LCDBIAS1 RE0 LCDBIAS1	2	I/O I	ST Analog	Digital I/O. BIAS1 input for LCD.
RE1/LCDBIAS2 RE1 LCDBIAS2	1	I/O I	ST Analog	Digital I/O. BIAS2 input for LCD.
LCDBIAS3	64	Т	Analog	BIAS3 input for LCD.
RE3/COM0 RE3 COM0	63	I/O O	ST Analog	Digital I/O. COM0 output for LCD.
RE4/COM1 RE4 COM1	62	I/O O	ST Analog	Digital I/O. COM1 output for LCD.
RE5/COM2 RE5 COM2	61	I/O O	ST Analog	Digital I/O. COM2 output for LCD.
RE6/COM3 RE6 COM3	60	I/O O	ST Analog	Digital I/O. COM3 output for LCD.
RE7/CCP2/SEG31 RE7 CCP2 ⁽²⁾ SEG31	59	I/O I/O O	ST ST Analog	Digital I/O. Capture 2 input/Compare 2 output/PWM 2 output. SEG31 output for LCD.
I = Input P = Power	itt Trigger input			CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

TABLE 1-3: PIC18F6XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)

Note 1: Default assignment for CCP2 when CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin	Buffer	Description
Pin Name	TQFP	Туре	Туре	Description
				PORTF is a bidirectional I/O port.
RF1/AN6/C2OUT/SEG19 RF1 AN6 C2OUT SEG19	17	I/O I O O	ST Analog — Analog	Digital I/O. Analog input 6. Comparator 2 output. SEG19 output for LCD.
RF2/AN7/C1OUT/SEG20 RF2 AN7 C1OUT SEG20	16	I/O I O O	ST Analog — Analog	Digital I/O. Analog input 7. Comparator 1 output. SEG20 output for LCD.
RF3/AN8/SEG21 RF3 AN8 SEG21	15	I/O I O	ST Analog Analog	Digital I/O. Analog input 8. SEG21 output for LCD.
RF4/AN9/SEG22 RF4 AN9 SEG22	14	I/O I O	ST Analog Analog	Digital I/O. Analog input 9. SEG22 output for LCD.
RF5/AN10/CVREF/SEG23 RF5 AN10 CVREF SEG23	13	I/O I O O	ST Analog Analog Analog	Digital I/O. Analog input 10. Comparator reference voltage output. SEG23 output for LCD.
RF6/AN11/SEG24 RF6 AN11 SEG24	12	I/O I O	ST Analog Analog	Digital I/O. Analog input 11. SEG24 output for LCD.
RF7/AN5/SS/SEG25 RF7 AN5 SS SEG25	11	I/O O I O	ST Analog TTL Analog	Digital I/O. Analog input 5. SPI slave select input. SEG25 output for LCD.
I = Input P = Power	t Trigger input v			CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

TABLE 1-3:	PIC18F6XJ90 PINOUT I/O DESCRIPTIONS ((CONTINUED)

Note 1: Default assignment for CCP2 when CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin	Buffer	Description	
Pin Name	TQFP	Туре	Туре	Description	
				PORTG is a bidirectional I/O port.	
RG0/LCDBIAS0 RG0 LCDBIAS0	3	I/O I	ST Analog	Digital I/O. BIAS0 input for LCD.	
RG1/TX2/CK2 RG1 TX2 CK2	4	I/O O I/O	ST — ST	Digital I/O. AUSART asynchronous transmit. AUSART synchronous clock (see related RX2/DT2).	
RG2/RX2/DT2/VLCAP1 RG2 RX2 DT2 VLCAP1	5	I/O I I/O I	ST ST ST Analog	Digital I/O. AUSART asynchronous receive. AUSART synchronous data (see related TX2/CK2). LCD charge pump capacitor input.	
RG3/VLCAP2 RG3 VLCAP2	6	I/O I	ST Analog	Digital I/O. LCD charge pump capacitor input.	
RG4/SEG26 RG4 SEG26	8	I/O O	ST Analog	Digital I/O. SEG26 output for LCD.	
Vss	9, 25, 41, 56	Р		Ground reference for logic and I/O pins.	
Vdd	26, 38, 57	Р		Positive supply for logic and I/O pins.	
AVss	20	Р		Ground reference for analog modules.	
AVDD	19	Р		Positive supply for analog modules.	
ENVREG	18	I	ST	Enable for on-chip voltage regulator.	
VDDCORE/VCAP VDDCORE	10	Ρ	—	Core logic power or external filter capacitor connection. Positive supply for microcontroller core logic (regulator disabled).	
VCAP		Р	—	External filter capacitor connection (regulator enabled).	
Legend:TTL= TTL compatible inputCMOS= CMOS compatible input or outputST= Schmitt Trigger input with CMOS levelsAnalog= Analog inputI= InputO= OutputP= PowerOD= Open-Drain (no P diode to VDD)					

TABLE 1-3: PIC18F6XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)

Note 1: Default assignment for CCP2 when CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin	Buffer	Description
	TQFP	Туре	Туре	
MCLR	9	Ι	ST	Master Clear (input) or programming voltage (input). This pin is an active-low Reset to the device.
OSC1/CLKI/RA7 OSC1 CLKI	49	I	CMOS CMOS	Oscillator crystal or external clock input. Oscillator crystal input. External clock source input. Always associated with pin function OSC1. (See related OSC1/CLKI, OSC2/CLKO pins.)
RA7		I/O	TTL	General purpose I/O pin.
OSC2/CLKO/RA6 OSC2	50	ο	_	Oscillator crystal or clock output. Oscillator crystal output. Connects to crystal or resonator in Crystal Oscillator mode.
CLKO		0	_	In EC modes, OSC2 pin outputs CLKO, which has 1/4 the frequency of OSC1 and denotes the instruction cycle rate.
RA6		I/O	TTL	General purpose I/O pin.
				PORTA is a bidirectional I/O port.
RA0/AN0 RA0 AN0	30	I/O I	TTL Analog	Digital I/O. Analog input 0.
RA1/AN1/SEG18 RA1 AN1 SEG18	29	I/O I O	TTL Analog Analog	Digital I/O. Analog input 1. SEG18 output for LCD.
RA2/AN2/VREF- RA2 AN2 VREF-	28	I/O I I	TTL Analog Analog	Digital I/O. Analog input 2. A/D reference voltage (Low) input.
RA3/AN3/VREF+ RA3 AN3 VREF+	27	I/O I I	TTL Analog Analog	Digital I/O. Analog input 3. A/D reference voltage (High) input.
RA4/T0CKI/SEG14 RA4 T0CKI SEG14	34	I/O I O	ST/OD ST Analog	Digital I/O. Open-drain when configured as output. Timer0 external clock input. SEG14 output for LCD.
RA5/AN4/SEG15 RA5 AN4 SEG15	33	I/O I O	TTL Analog Analog	Digital I/O. Analog input 4. SEG15 output for LCD.
RA6				See the OSC2/CLKO/RA6 pin.
RA7				See the OSC1/CLKI/RA7 pin.
ST = Schrr I = Input P = Powe	er			CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

TABLE 1-4:	PIC18F8XJ90 PINOUT I/O DESCRIPTIONS

Note 1: Default assignment for CCP2 when CCP2MX Configuration bit is set.

Bin Nomo	Pin Number	Pin	in Buffer	Description
Pin Name	TQFP	Туре	Туре	Description
				PORTB is a bidirectional I/O port. PORTB can be software programmed for internal weak pull-ups on all inputs.
RB0/INT0/SEG30 RB0 INT0 SEG30	58	I/O I O	TTL ST Analog	Digital I/O. External interrupt 0. SEG30 output for LCD.
RB1/INT1/SEG8 RB1 INT1 SEG8	57	I/O I O	TTL ST Analog	Digital I/O. External interrupt 1. SEG8 output for LCD.
RB2/INT2/SEG9 RB2 INT2 SEG9	56	I/O I O	TTL ST Analog	Digital I/O. External interrupt 2. SEG9 output for LCD.
RB3/INT3/SEG10 RB3 INT3 SEG10	55	I/O I O	TTL ST Analog	Digital I/O. External interrupt 3. SEG10 output for LCD.
RB4/KBI0/SEG11 RB4 KBI0 SEG11	54	I/O I O	TTL TTL Analog	Digital I/O. Interrupt-on-change pin. SEG11 output for LCD.
RB5/KBI1/SEG29 RB5 KBI1 SEG29	53	I/O I O	TTL TTL Analog	Digital I/O. Interrupt-on-change pin. SEG29 output for LCD.
RB6/KBI2/PGC RB6 KBI2 PGC	52	I/O I I/O	TTL TTL ST	Digital I/O. Interrupt-on-change pin. In-Circuit Debugger and ICSP™ programming clock pin.
RB7/KBI3/PGD RB7 KBI3 PGD	47	I/O I I/O	TTL TTL ST	Digital I/O. Interrupt-on-change pin. In-Circuit Debugger and ICSP programming data pin.
			MOS leve	CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

TABLE 1-4: PIC18F8XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)

Note 1: Default assignment for CCP2 when CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin	Buffer	Description
	TQFP	Туре Туре		Description
				PORTC is a bidirectional I/O port.
RC0/T10SO/T13CKI RC0 T10SO T13CKI	36	I/O O I	ST — ST	Digital I/O. Timer1 oscillator output. Timer1/Timer3 external clock input.
RC1/T1OSI/CCP2/SEG32 RC1 T1OSI CCP2 ⁽¹⁾ SEG32	35	I/O I I/O O	ST CMOS ST Analog	Digital I/O. Timer1 oscillator input. Capture2 input/Compare2 output/PWM2 output. SEG32 output for LCD.
RC2/CCP1/SEG13 RC2 CCP1 SEG13	43	I/O I/O O	ST ST Analog	Digital I/O. Capture1 input/Compare1 output/PWM1 output. SEG13 output for LCD.
RC3/SCK/SCL/SEG17 RC3 SCK SCL SEG17	44	I/O I/O I/O O	ST ST ST Analog	Digital I/O. Synchronous serial clock input/output for SPI mode. Synchronous serial clock input/output for I ² C™ mode. SEG17 output for LCD.
RC4/SDI/SDA/SEG16 RC4 SDI SDA SEG16	45	I/O I I/O O	ST ST ST Analog	Digital I/O. SPI data in. I ² C data I/O. SEG16 output for LCD.
RC5/SDO/SEG12 RC5 SDO SEG12	46	I/O O O	ST — Analog	Digital I/O. SPI data out. SEG12 output for LCD.
RC6/TX1/CK1/SEG27 RC6 TX1 CK1 SEG27	37	I/O O I/O O	ST — ST Analog	Digital I/O. EUSART asynchronous transmit. EUSART synchronous clock (see related RX1/DT1). SEG27 output for LCD.
RC7/RX1/DT1/SEG28 RC7 RX1 DT1 SEG28	38	I/O I I/O O	ST ST ST Analog	Digital I/O. EUSART asynchronous receive. EUSART synchronous data (see related TX1/CK1). SEG28 output for LCD.
Legend: TTL = TTL compatible input CMOS = CMOS compatible input or output ST = Schmitt Trigger input with CMOS levels Analog = Analog input I = Input O = Output P = Power OD = Open-Drain (no P diode to VDD)				

TABLE 1-4: PIC18F8XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)

Pin Name	Pin Number	Pin	Buffer	Description
Pin Name	TQFP	Туре	Туре	Description
				PORTD is a bidirectional I/O port.
RD0/SEG0 RD0 SEG0	72	I/O O	ST Analog	Digital I/O. SEG0 output for LCD.
RD1/SEG1 RD1 SEG1	69	I/O O	ST Analog	Digital I/O. SEG1 output for LCD.
RD2/SEG2 RD2 SEG2	68	I/O O	ST Analog	Digital I/O. SEG2 output for LCD.
RD3/SEG3 RD3 SEG3	67	I/O O	ST Analog	Digital I/O. SEG3 output for LCD.
RD4/SEG4 RD4 SEG4	66	I/O O	ST Analog	Digital I/O. SEG4 output for LCD.
RD5/SEG5 RD5 SEG5	65	I/O O	ST Analog	Digital I/O. SEG5 output for LCD.
RD6/SEG6 RD6 SEG6	64	I/O O	ST Analog	Digital I/O. SEG6 output for LCD.
RD7/SEG7 RD7 SEG7	63	I/O O	ST Analog	Digital I/O. SEG7 output for LCD.
Legend: TTL = TTL co ST = Schmi I = Input	ompatible input tt Trigger input	with C	MOS leve	CMOS = CMOS compatible input or output Analog = Analog input O = Output

TABLE 1-4: PIC18F8XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)

= Input Р

Output

= Power

OD = Open-Drain (no P diode to VDD)

Note 1: Default assignment for CCP2 when CCP2MX Configuration bit is set.

Dia Maraa	Pin Number	Pin	Buffer	Description
Pin Name	TQFP	Туре	Туре	Description
				PORTE is a bidirectional I/O port.
RE0/LCDBIAS1 RE0 LCDBIAS1	4	I/O I	ST Analog	Digital I/O. BIAS1 input for LCD.
RE1/LCDBIAS2 RE1 LCDBIAS2	3	I/O I	ST Analog	Digital I/O. BIAS2 input for LCD.
LCDBIAS3	78	I	Analog	BIAS3 input for LCD.
RE3/COM0 RE3 COM0	77	I/O O	ST Analog	Digital I/O. COM0 output for LCD.
RE4/COM1 RE4 COM1	76	I/O O	ST Analog	Digital I/O. COM1 output for LCD.
RE5/COM2 RE5 COM2	75	I/O O	ST Analog	Digital I/O. COM2 output for LCD.
RE6/COM3 RE6 COM3	74	I/O O	ST Analog	Digital I/O. COM3 output for LCD.
RE7/CCP2/SEG31 RE7 CCP2 ⁽²⁾ SEG31	73	I/O I/O O	ST ST Analog	Digital I/O. Capture 2 input/Compare 2 output/PWM 2 output. SEG31 output for LCD.
I = Input P = Powe	itt Trigger input r	with C		CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

TABLE 1-4: PIC18F8XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)

Note 1: Default assignment for CCP2 when CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin	Buffer	Description
Pin Name	TQFP	Туре	Туре	Description
				PORTF is a bidirectional I/O port.
RF1/AN6/C2OUT/SEG19 RF1 AN6 C2OUT SEG19	23	I/O I O O	ST Analog — Analog	Digital I/O. Analog input 6. Comparator 2 output. SEG19 output for LCD.
RF2/AN7/C1OUT/SEG20 RF2 AN7 C1OUT SEG20	18	I/O I O O	ST Analog — Analog	Digital I/O. Analog input 7. Comparator 1 output. SEG20 output for LCD.
RF3/AN8/SEG21 RF3 AN8 SEG21	17	I/O I O	ST Analog Analog	Digital I/O. Analog input 8. SEG21 output for LCD.
RF4/AN9/SEG22 RF4 AN9 SEG22	16	I/O I O	ST Analog Analog	Digital I/O. Analog input 9. SEG22 output for LCD.
RF5/AN10/CVREF/SEG23 RF5 AN10 CVREF SEG23	15	I/O I O O	ST Analog Analog Analog	Digital I/O. Analog input 10. Comparator reference voltage output. SEG23 output for LCD.
RF6/AN11/SEG24 RF6 AN11 SEG24	14	I/O I O	ST Analog Analog	Digital I/O. Analog input 11. SEG24 output for LCD.
RF7/AN5/SS/SEG25 RF7 AN5 SS SEG25	13	I/O O I O	ST Analog TTL Analog	Digital I/O. Analog input 5. SPI slave select input. SEG25 output for LCD.
Legend: TTL = TTL co ST = Schmi I = Input P = Power	tt Trigger input	with C	MOS leve	CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

TABLE 1-4: PIC18F8XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)

Note 1: Default assignment for CCP2 when CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin	Buffer	Description	
Pin Name	TQFP	Туре	Туре	Description	
				PORTG is a bidirectional I/O port.	
RG0/LCDBIAS0 RG0 LCDBIAS0	5	I/O I	ST Analog	Digital I/O. BIAS0 input for LCD.	
RG1/TX2/CK2 RG1 TX2 CK2	6	I/O O I/O	ST — ST	Digital I/O. AUSART asynchronous transmit. AUSART synchronous clock (see related RX2/DT2).	
RG2/RX2/DT2/VLCAP1 RG2 RX2 DT2 VLCAP1	7	I/O /O 	ST ST ST Analog	Digital I/O. AUSART asynchronous receive. AUSART synchronous data (see related TX2/CK2). LCD charge pump capacitor input.	
RG3/VLCAP2 RG3 VLCAP2	8	I/O I	ST Analog	Digital I/O. LCD charge pump capacitor input.	
RG4/SEG26 RG4 SEG26	10	I/O O	ST Analog	Digital I/O. SEG26 output for LCD.	
Legend: TTL = TTL co ST = Schmi I = Input P = Power	tt Trigger input			O = Output OD = Open-Drain (no P diode to VDD)	

TABLE 1-4:	PIC18F8XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)	

Note 1: Default assignment for CCP2 when CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin	Buffer	Description	
Pin Name	TQFP	Туре	Туре	Description	
				PORTH is a bidirectional I/O port.	
RH0/SEG47 RH0 SEG47	79	I/O O	ST Analog	Digital I/O. SEG47 output for LCD.	
RH1/SEG46 RH1 SEG46	80	I/O O	ST Analog	Digital I/O. SEG46 output for LCD.	
RH2/SEG45 RH2 SEG45	1	I/O O	ST Analog	Digital I/O. SEG45 output for LCD.	
RH3/SEG44 RH3 SEG44	2	I/O O	ST Analog	Digital I/O. SEG44 output for LCD.	
RH4/SEG40 RH4 SEG40	22	I/O O	ST Analog	Digital I/O. SEG40 output for LCD.	
RH5/SEG41 RH5 SEG41	21	I/O O	ST Analog	Digital I/O. SEG41 output for LCD.	
RH6/SEG42 RH6 SEG42	20	I/O O	ST Analog	Digital I/O. SEG42 output for LCD.	
RH7/SEG43 RH7 SEG43	19	I/O O	ST Analog	Digital I/O. SEG43 output for LCD.	
Legend: TTL = TTL compatible input CMOS = CMOS compatible input or output ST = Schmitt Trigger input with CMOS levels Analog = Analog input I = Input O = Output					

TABLE 1-4: PIC18F8XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)

Р = Power = Open-Drain (no P diode to VDD)

OD

Note 1: Default assignment for CCP2 when CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin	Buffer	Description	
Pin Name	TQFP	Туре	Туре	Description	
				PORTJ is a bidirectional I/O port.	
RJ0	62	I/O	ST	Digital I/O.	
RJ1/SEG33 RJ1 SEG33	61	I/O O	ST Analog	Digital I/O. SEG33 output for LCD.	
RJ2/SEG34 RJ2 SEG34	60	I/O O	ST Analog	Digital I/O. SEG34 output for LCD.	
RJ3/SEG35 RJ3 SEG35	59	I/O O	ST Analog	Digital I/O. SEG35 output for LCD.	
RJ4/SEG39 RJ4 SEG39	39	I/O O	ST Analog	Digital I/O. SEG39 output for LCD.	
RJ5/SEG38 RJ5 SEG38	40	I/O O	ST Analog	Digital I/O SEG38 output for LCD.	
RJ6/SEG37 RJ6 SEG37	41	I/O O	ST Analog	Digital I/O. SEG37 output for LCD.	
RJ7/SEG36 RJ7 SEG36	42	I/O O	ST Analog	Digital I/O. SEG36 output for LCD.	
Vss	11, 31, 51, 70	Р		Ground reference for logic and I/O pins.	
Vdd	32, 48, 71	Р		Positive supply for logic and I/O pins.	
AVss	26	Р		Ground reference for analog modules.	
AVDD	25	Р	—	Positive supply for analog modules.	
ENVREG	24	Ι	ST	Enable for on-chip voltage regulator.	
VDDCORE/VCAP VDDCORE	12	Ρ	_	Core logic power or external filter capacitor connection. Positive supply for microcontroller core logic (regulator disabled).	
VCAP P (regulator disabled). External filter capacitor connection (regulator enabled). Legend: TTL = TTL compatible input ST = Schmitt Trigger input with CMOS levels I = Input P = Power CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD) Note 1: Default assignment for CCP2 when CCP2MX Configuration bit is set.					

TABLE 1-4:	PIC18F8XJ90 PINOUT I/O DESCRIPTIONS ((CONTINUED))

NOTES:

2.0 OSCILLATOR CONFIGURATIONS

2.1 Oscillator Types

The PIC18F85J90 family of devices can be operated in six different oscillator modes:

- 1. HS High-Speed Crystal/Resonator
- 2. HSPLL High-Speed Crystal/Resonator with Software PLL Control
- 3. EC External Clock with Fosc/4 Output
- 4. ECPLL External Clock with Software PLL Control
- 5. INTOSC Internal Fast RC (8 MHz) oscillator
- 6. INTRC Internal 31 kHz Oscillator

Five of these are selected by the user by programming the FOSC2:FOSC0 Configuration bits. The sixth mode (INTRC) may be invoked under software control; it can also be configured as the default mode on device Resets.

In addition, PIC18F85J90 family devices can switch between different clock sources, either under software control, or automatically under certain conditions. This allows for additional power savings by managing device clock speed in real time without resetting the application.

The clock sources for the PIC18F85J90 family of devices are shown in Figure 2-1.

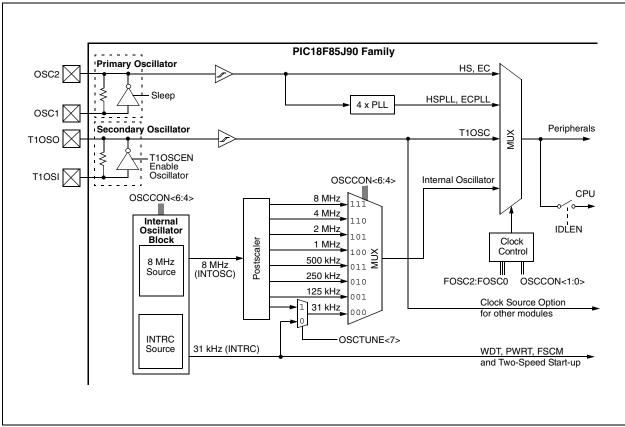


FIGURE 2-1: PIC18F85J90 FAMILY CLOCK DIAGRAM

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2.2 Control Registers

The OSCCON register (Register 2-1) controls the main aspects of the device clock's operation. It selects the oscillator type to be used, which of the power-managed modes to invoke and the output frequency of the INTOSC source. It also provides status on the oscillators. The OSCTUNE register (Register 2-2) controls the tuning and operation of the internal oscillator block. It also implements the PLLEN bits, which control the operation of the Phase Locked Loop (PLL) in Internal Oscillator modes (see Section 2.4.3 "PLL Frequency Multiplier").

REGISTER 2-1: OSCCON: OSCILLATOR CONTROL REGISTER

R/W-0	R/W-1	R/W-0	R/W-0	R ⁽¹⁾	R-0	R/W-0	R/W-0			
IDLEN	IRCF2	IRCF1	IRCF0	OSTS	IOFS	SCS1	SCS0			
bit 7							bit 0			
Legend: R = Reada	abla bit	W = Writable	hit	U = Unimpler	ponted bit rea	d ac '0'				
-n = Value		1' = Bit is set		$0^{\circ} = \text{Bit is clear}$		x = Bit is unkr				
bit 7	IDLEN: Idle I	Enable bit								
	1 = Device e	1 = Device enters an Idle mode when a SLEEP instruction is executed								
		•		EEP instruction	is executed					
bit 6-4		0: INTOSC Sou								
	$111 = 8 \text{ MH}_{110} = 4 \text{ MH}_{110}$	z (INTOSC driv	es clock direct	ly)						
	101 = 2 MHz	_								
	100 = 1 MHz									
	011 = 500 k	011 = 500 kHz								
		010 = 250 kHz								
	001 = 125 k		NTOSC/256 or							
bit 3		000 = 31 kHz (from either INTOSC/256 or INTRC) ⁽³⁾ OSTS: Oscillator Start-up Time-out Status bit ⁽¹⁾								
bit 0	1 = Oscillator Start-up Timer (OST) time-out has expired; primary oscillator is running									
			. ,	out is running; pr	• •	•				
bit 2	IOFS: INTOS	SC Frequency S	table bit		-	-				
	1 = Fast RC	oscillator frequ	ency is stable							
	0 = Fast RC	oscillator frequ	ency is not sta	ble						
bit 1-0	SCS1:SCS0	System Clock	Select bits ⁽⁴⁾							
		oscillator block								
	10 = Primary oscillator									
	01 = Timer1									
	<u>When FOSC</u> 00 = Primary									
	When FOSC									
	00 = Internal									
Note 1:	Reset state depe	ends on state of	the IESO Con	figuration bit.						
2:	Modifying these I			•	switch if the in	ternal oscillator	is providina			

- 2: Modifying these bits will cause an immediate clock frequency switch if the internal oscillator is providing the device clocks.
- **3:** Source selected by the INTSRC bit (OSCTUNE<7>), see text.
- 4: Modifying these bits will cause an immediate clock source switch.

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	
INTSRC	PLLEN ⁽¹⁾	TUN5	TUN4	TUN3	TUN2	TUN1	TUN0	
bit 7							bit 0	
Legend:								
R = Readable	e bit	W = Writable	bit	U = Unimplem	nented bit, read	1 as '0'		
-n = Value at	POR	'1' = Bit is set		•		x = Bit is unkn	Bit is unknown	
bit 6	 1 = 31.25 kHz device clock derived from 8 MHz INTOSC source (divide-by-256 enabled) 0 = 31 kHz device clock derived from INTRC 31 kHz oscillator PLLEN: Frequency Multiplier PLL Enable bit⁽¹⁾ 1 = PLL enabled 							
bit 5-0	 0 = PLL disabled TUN5:TUN0: Fast RC Oscillator (INTOSC) Frequency Tuning bits 011111 = Maximum frequency • 							
	• 000001 000000 = Ce 111111 •	• nter frequency. •	Fast RC osci	llator is running	at the calibrate	ed frequency.		

REGISTER 2-2: OSCTUNE: OSCILLATOR TUNING REGISTER

Note 1: Available only for ECPLL and HSPLL oscillator configurations; otherwise, this bit is unavailable and reads as '0'.

2.3 Clock Sources and Oscillator Switching

Essentially, PIC18F85J90 family devices have three independent clock sources:

- · Primary oscillators
- · Secondary oscillators
- Internal oscillator

The **primary oscillators** can be thought of as the main device oscillators. These are any external oscillators connected to the OSC1 and OSC2 pins, and include the External Crystal and Resonator modes and the External Clock modes. In some circumstances, the internal oscillator block may be considered a primary oscillator. The particular mode is defined by the FOSC Configuration bits. The details of these modes are covered in **Section 2.4 "External Oscillator Modes**".

The **secondary oscillators** are external clock sources that are not connected to the OSC1 or OSC2 pins. These sources may continue to operate even after the controller is placed in a power-managed mode. PIC18F85J90 family devices offer the Timer1 oscillator as a secondary oscillator source. This oscillator, in all power-managed modes, is often the time base for functions such as a Real-Time Clock. The Timer1 oscillator is discussed in greater detail in **Section 11.3** "**Timer1 Oscillator**"

In addition to being a primary clock source in some circumstances, the **internal oscillator** is available as a power-managed mode clock source. The INTRC source is also used as the clock source for several special features, such as the WDT and Fail-Safe Clock Monitor. The internal oscillator block is discussed in more detail in **Section 2.5** "Internal Oscillator **Block**".

The PIC18F85J90 family includes features that allow the device clock source to be switched from the main oscillator, chosen by device configuration, to one of the alternate clock sources. When an alternate clock source is enabled, various power-managed operating modes are available.

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2.3.1 CLOCK SOURCE SELECTION

The System Clock Select bits, SCS1:SCS0 (OSCCON<1:0>), select the clock source. The available clock sources are the primary clock defined by the FOSC1:FOSC0 Configuration bits, the secondary clock (Timer1 oscillator) and the internal oscillator. The clock source changes after one or more of the bits are written to, following a brief clock transition interval.

The OSTS (OSCCON<3>) and T1RUN (T1CON<6>) bits indicate which clock source is currently providing the device clock. The OSTS bit indicates that the Oscillator Start-up Timer (OST) has timed out and the primary clock is providing the device clock in primary clock modes. The T1RUN bit indicates when the Timer1 oscillator is providing the device clock in secondary clock modes. In power-managed modes, only one of these bits will be set at any time. If neither of these bits are set, the INTRC is providing the clock, or the internal oscillator has just started and is not yet stable.

The IDLEN bit determines if the device goes into Sleep mode or one of the Idle modes when the SLEEP instruction is executed.

The use of the flag and control bits in the OSCCON register is discussed in more detail in **Section 3.0 "Power-Managed Modes"**.

- Note 1: The Timer1 oscillator must be enabled to select the secondary clock source. The Timer1 oscillator is enabled by setting the T1OSCEN bit in the Timer1 Control register (T1CON<3>). If the Timer1 oscillator is not enabled, then any attempt to select a secondary clock source when executing a SLEEP instruction will be ignored.
 - 2: It is recommended that the Timer1 oscillator be operating and stable before executing the SLEEP instruction or a very long delay may occur while the Timer1 oscillator starts.

2.3.1.1 System Clock Selection and the FOSC2 Configuration Bit

The SCS bits are cleared on all forms of Reset. In the device's default configuration, this means the primary oscillator defined by FOSC1:FOSC0 (that is, one of the HS or EC modes) is used as the primary clock source on device Resets.

The default clock configuration on Reset can be changed with the FOSC2 Configuration bit. This bit determines whether the external or internal oscillator will be the default clock source on subsequent device Resets. By extension, it also has the effect of determining the clock source selected when SCS1:SCS0 are in their Reset state (= 00). When FOSC2 = 1 (default), the oscillator source defined by FOSC1:FOSC0 is selected whenever SCS1:SCS0 = 00. When FOSC2 = 0, the internal oscillator block is selected whenever SCS1:SCS2 = 00.

In those cases when the internal oscillator block is the default clock on Reset, the Fast RC oscillator (INTOSC) will be used as the device clock source. It will initially start at 1 MHz, the postscaler selection that corresponds to the Reset value of the IRCF2:IRCF0 bits ('100').

Regardless of the setting of FOSC2, INTRC will always be enabled on device power-up. It serves as the clock source until the device has loaded its configuration values from memory. It is at this point that the FOSC Configuration bits are read and the oscillator selection of the operational mode is made.

Note that either the primary clock or the internal oscillator will have two bit setting options for the possible values of SCS1:SCS0, at any given time, depending on the setting of FOSC2.

2.3.2 OSCILLATOR TRANSITIONS

PIC18F85J90 family devices contain circuitry to prevent clock "glitches" when switching between clock sources. A short pause in the device clock occurs during the clock switch. The length of this pause is the sum of two cycles of the old clock source and three to four cycles of the new clock source. This formula assumes that the new clock source is stable.

Clock transitions are discussed in greater detail in **Section 3.1.2 "Entering Power-Managed Modes"**.

2.4 External Oscillator Modes

2.4.1 CRYSTAL OSCILLATOR/CERAMIC RESONATORS (HS MODES)

In HS or HSPLL Oscillator modes, a crystal or ceramic resonator is connected to the OSC1 and OSC2 pins to establish oscillation. Figure 2-2 shows the pin connections.

The oscillator design requires the use of a parallel cut crystal.

Note: Use of a series cut crystal may give a frequency out of the crystal manufacturer's specifications.

TABLE 2-1:CAPACITOR SELECTION FOR
CERAMIC RESONATORS

Typical Capacitor Values Used:					
Mode	Freq.	OSC1	OSC2		
HS	8.0 MHz 16.0 MHz	27 pF 22 pF	27 pF 22 pF		

Capacitor values are for design guidance only.

Different capacitor values may be required to produce acceptable oscillator operation. The user should test the performance of the oscillator over the expected VDD and temperature range for the application. Refer to the following application notes for oscillator specific information:

- AN588, "PIC[®] Microcontroller Oscillator Design Guide"
- AN826, "Crystal Oscillator Basics and Crystal Selection for rfPIC[®] and PIC[®] Devices"
- AN849, 'Basic PIC[®] Oscillator Design"
- AN943, "Practical PIC[®] Oscillator Analysis and Design"
- AN949, "Making Your Oscillator Work"

See the notes following Table 2-2 for additional information.

TABLE 2-2:CAPACITOR SELECTION FOR
CRYSTAL OSCILLATOR

Osc Type	Crystal Freq.	Typical Capacitor Values Tested:		
	Fieq.	C1	C2	
HS	4 MHz	27 pF	27 pF	
	8 MHz	22 pF	22 pF	
	20 MHz	15 pF	15 pF	

Capacitor values are for design guidance only.

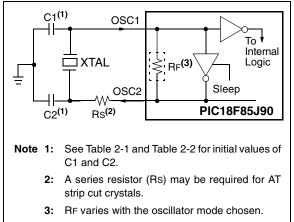
Different capacitor values may be required to produce acceptable oscillator operation. The user should test the performance of the oscillator over the expected VDD and temperature range for the application.

Refer to the Microchip application notes cited in Table 2-1 for oscillator specific information. Also see the notes following this table for additional information.

- Note 1: Higher capacitance increases the stability of oscillator but also increases the start-up time.
 - 2: Since each resonator/crystal has its own characteristics, the user should consult the resonator/crystal manufacturer for appropriate values of external components.
 - **3:** Rs may be required to avoid overdriving crystals with low drive level specification.
 - 4: Always verify oscillator performance over the VDD and temperature range that is expected for the application.

FIGURE 2-2:

CRYSTAL/CERAMIC RESONATOR OPERATION (HS OR HSPLL CONFIGURATION)

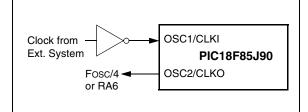


2.4.2 EXTERNAL CLOCK INPUT (EC MODES)

The EC and ECPLL Oscillator modes require an external clock source to be connected to the OSC1 pin. There is no oscillator start-up time required after a Power-on Reset or after an exit from Sleep mode.

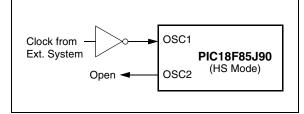
In the EC Oscillator mode, the oscillator frequency divided by 4 is available on the OSC2 pin. This signal may be used for test purposes or to synchronize other logic. Figure 2-3 shows the pin connections for the EC Oscillator mode.

FIGURE 2-3: EXTERNAL CLOCK INPUT OPERATION (EC CONFIGURATION)



An external clock source may also be connected to the OSC1 pin in the HS mode, as shown in Figure 2-4. In this configuration, the divide-by-4 output on OSC2 is not available.

FIGURE 2-4: EXTERNAL CLOCK INPUT OPERATION (HS OSC CONFIGURATION)

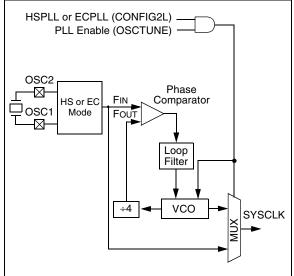


2.4.3 PLL FREQUENCY MULTIPLIER

A Phase Locked Loop (PLL) circuit is provided as an option for users who want to use a lower frequency oscillator circuit, or to clock the device up to its highest rated frequency from a crystal oscillator. This may be useful for customers who are concerned with EMI due to high-frequency crystals, or users who require higher clock speeds from an internal oscillator. For these reasons, the HSPLL and ECPLL modes are available.

The HSPLL and ECPLL modes provide the ability to selectively run the device at 4 times the external oscillating source to produce frequencies up to 40 MHz. The PLL is enabled by programming the FOSC2:FOSC0 Configuration bits (CONFIG2L<2:0>) to either '110' (for ECPLL) or '100' (for HSPLL). In addition, the PLLEN bit (OSCTUNE<6>) must also be set. Clearing PLLEN disables the PLL, regardless of the chosen oscillator configuration. It also allows additional flexibility for controlling the application's clock speed in software.





2.5 Internal Oscillator Block

The PIC18F85J90 family of devices includes an internal oscillator block which generates two different clock signals; either can be used as the microcontroller's clock source. This may eliminate the need for an external oscillator circuit on the OSC1 and/or OSC2 pins.

The main output is the Fast RC oscillator, or INTOSC, an 8 MHz clock source which can be used to directly drive the device clock. It also drives a postscaler, which can provide a range of clock frequencies from 31 kHz to 4 MHz. INTOSC is enabled when a clock frequency from 125 kHz to 8 MHz is selected. The INTOSC output can also be enabled when 31 kHz is selected, depending on the INTSRC bit (OSCTUNE<7>).

The other clock source is the Internal RC oscillator (INTRC), which provides a nominal 31 kHz output. INTRC is enabled if it is selected as the device clock source. It is also enabled automatically when any of the following are enabled:

- Power-up Timer
- Fail-Safe Clock Monitor
- Watchdog Timer
- Two-Speed Start-up

These features are discussed in greater detail in Section 22.0 "Special Features of the CPU".

The clock source frequency (INTOSC direct, INTOSC with postscaler or INTRC direct) is selected by configuring the IRCF bits of the OSCCON register. The default frequency on device Resets is 1 MHz.

2.5.1 OSC1 AND OSC2 PIN CONFIGURATION

Whenever the internal oscillator is configured as the default clock source (FOSC2 = 0), the OSC1 and OSC2 pins are reconfigured automatically as port pins, RA6 and RA7. In this mode, they function as general digital I/O. All oscillator functions on the pins are disabled.

2.5.2 INTERNAL OSCILLATOR OUTPUT FREQUENCY AND TUNING

The internal oscillator block is calibrated at the factory to produce an INTOSC output frequency of 8 MHz. It can be adjusted in the user's application by writing to TUN5:TUN0 (OSCTUNE<5:0>) in the OSCTUNE register (Register 2-2).

When the OSCTUNE register is modified, the INTOSC frequency will begin shifting to the new frequency. The oscillator will stabilize within 1 ms. Code execution continues during this shift. There is no indication that the shift has occurred.

The INTRC oscillator operates independently of the INTOSC source. Any changes in INTOSC across voltage and temperature are not necessarily reflected by changes in INTRC or vice versa. The frequency of INTRC is not affected by OSCTUNE.

2.5.3 INTOSC FREQUENCY DRIFT

The INTOSC frequency may drift as VDD or temperature changes, and can affect the controller operation in a variety of ways. It is possible to adjust the INTOSC frequency by modifying the value in the OSCTUNE register. This will have no effect on the INTRC clock source frequency.

Tuning INTOSC requires knowing when to make the adjustment, in which direction it should be made and in some cases, how large a change is needed. Three compensation techniques are shown here.

2.5.3.1 Compensating with the EUSART

An adjustment may be required when the EUSART begins to generate framing errors or receives data with errors while in Asynchronous mode. Framing errors indicate that the device clock frequency is too high. To adjust for this, decrement the value in OSCTUNE to reduce the clock frequency. On the other hand, errors in data may suggest that the clock speed is too low. To compensate, increment OSCTUNE to increase the clock frequency.

2.5.3.2 Compensating with the Timers

This technique compares device clock speed to some reference clock. Two timers may be used; one timer is clocked by the peripheral clock, while the other is clocked by a fixed reference source, such as the Timer1 oscillator.

Both timers are cleared, but the timer clocked by the reference generates interrupts. When an interrupt occurs, the internally clocked timer is read and both timers are cleared. If the internally clocked timer value is much greater than expected, then the internal oscillator block is running too fast. To adjust for this, decrement the OSCTUNE register.

2.5.3.3 Compensating with the CCP Module in Capture Mode

A CCP module can use free-running Timer1 (or Timer3), clocked by the internal oscillator block and an external event with a known period (i.e., AC power frequency). The time of the first event is captured in the CCPRxH:CCPRxL registers and is recorded for use later. When the second event causes a capture, the time of the first event is subtracted from the time of the second event. Since the period of the external event is known, the time difference between events can be calculated.

If the measured time is much greater than the calculated time, the internal oscillator block is running too fast. To compensate, decrement the OSCTUNE register. If the measured time is much less than the calculated time, the internal oscillator block is running too slow. To compensate, increment the OSCTUNE register.

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2.6 Effects of Power-Managed Modes on the Various Clock Sources

When PRI_IDLE mode is selected, the designated primary oscillator continues to run without interruption. For all other power-managed modes, the oscillator using the OSC1 pin is disabled. The OSC1 pin (and OSC2 pin if used by the oscillator) will stop oscillating.

In secondary clock modes (SEC_RUN and SEC_IDLE), the Timer1 oscillator is operating and providing the device clock. The Timer1 oscillator may also run in all power-managed modes if required to clock Timer1 or Timer3.

In RC_RUN and RC_IDLE modes, the internal oscillator provides the device clock source. The 31 kHz INTRC output can be used directly to provide the clock and may be enabled to support various special features, regardless of the power-managed mode (see Section 22.2 "Watchdog Timer (WDT)" through Section 22.5 "Fail-Safe Clock Monitor" for more information on WDT, Fail-Safe Clock Monitor and Two-Speed Start-up).

If the Sleep mode is selected, all clock sources are stopped. Since all the transistor switching currents have been stopped, Sleep mode achieves the lowest current consumption of the device (only leakage currents).

Enabling any on-chip feature that will operate during Sleep will increase the current consumed during Sleep. The INTRC is required to support WDT operation. The Timer1 oscillator may be operating to support a Real-Time Clock. Other features may be operating that do not require a device clock source (i.e., MSSP slave, PSP, INTn pins and others). Peripherals that may add significant current consumption are listed in Section 25.2 "DC Characteristics: Power-Down and Supply Current".

2.7 Power-up Delays

Power-up delays are controlled by two timers, so that no external Reset circuitry is required for most applications. The delays ensure that the device is kept in Reset until the device power supply is stable under normal circumstances and the primary clock is operating and stable. For additional information on power-up delays, see **Section 4.5 "Power-up Timer (PWRT)**".

The first timer is the Power-up Timer (PWRT), which provides a fixed delay on power-up (parameter 33, Table 25-11). It is always enabled.

The second timer is the Oscillator Start-up Timer (OST), intended to keep the chip in Reset until the crystal oscillator is stable (HS modes). The OST does this by counting 1024 oscillator cycles before allowing the oscillator to clock the device.

There is a delay of interval TCSD (parameter 38, Table 25-11), following POR, while the controller becomes ready to execute instructions.

Oscillator Mode	OSC1 Pin	OSC2 Pin
EC, ECPLL	Floating, pulled by external clock	At logic low (clock/4 output)
HS, HSPLL	Feedback inverter disabled at quiescent voltage level	Feedback inverter disabled at quiescent voltage level
INTOSC	I/O pin RA6, direction controlled by TRISA<6>	I/O pin RA7, direction controlled by TRISA<7>

 TABLE 2-3:
 OSC1 AND OSC2 PIN STATES IN SLEEP MODE

Note: See Table 4-2 in Section 4.0 "Reset" for time-outs due to Sleep and MCLR Reset.

3.0 POWER-MANAGED MODES

The PIC18F85J90 family devices provide the ability to manage power consumption by simply managing clocking to the CPU and the peripherals. In general, a lower clock frequency and a reduction in the number of circuits being clocked constitutes lower consumed power. For the sake of managing power in an application, there are three primary modes of operation:

- Run mode
- Idle mode
- Sleep mode

These modes define which portions of the device are clocked and at what speed. The Run and Idle modes may use any of the three available clock sources (primary, secondary or internal oscillator block); the Sleep mode does not use a clock source.

The power-managed modes include several power-saving features offered on previous PIC[®] devices. One is the clock switching feature, offered in other PIC18 devices, allowing the controller to use the Timer1 oscillator in place of the primary oscillator. Also included is the Sleep mode, offered by all PIC devices, where all device clocks are stopped.

3.1 Selecting Power-Managed Modes

Selecting a power-managed mode requires two decisions: if the CPU is to be clocked or not and which clock source is to be used. The IDLEN bit (OSCCON<7>) controls CPU clocking, while the SCS1:SCS0 bits (OSCCON<1:0>) select the clock source. The individual modes, bit settings, clock sources and affected modules are summarized in Table 3-1.

3.1.1 CLOCK SOURCES

The SCS1:SCS0 bits allow the selection of one of three clock sources for power-managed modes. They are:

- the primary clock, as defined by the FOSC2:FOSC0 Configuration bits
- the secondary clock (Timer1 oscillator)
- the internal oscillator

3.1.2 ENTERING POWER-MANAGED MODES

Switching from one power-managed mode to another begins by loading the OSCCON register. The SCS1:SCS0 bits select the clock source and determine which Run or Idle mode is to be used. Changing these bits causes an immediate switch to the new clock source, assuming that it is running. The switch may also be subject to clock transition delays. These are discussed in **Section 3.1.3 "Clock Transitions and Status Indicators"** and subsequent sections.

Entry to the power-managed Idle or Sleep modes is triggered by the execution of a SLEEP instruction. The actual mode that results depends on the status of the IDLEN bit.

Depending on the current mode and the mode being switched to, a change to a power-managed mode does not always require setting all of these bits. Many transitions may be done by changing the oscillator select bits, or changing the IDLEN bit, prior to issuing a SLEEP instruction. If the IDLEN bit is already configured correctly, it may only be necessary to perform a SLEEP instruction to switch to the desired mode.

Mada	oso	OSCCON bits		e Clocking	Ausilable Cleak and Casillater Courses				
Mode	IDLEN<7>(1)	IDLEN<7> ⁽¹⁾ SCS1:SCS0<1:0>		Peripherals	Available Clock and Oscillator Source				
Sleep	0	N/A	Off	Off	None – All clocks are disabled				
PRI_RUN	N/A	10	Clocked	Clocked	Primary – HS, EC, HSPLL, ECPLL; this is the normal full power execution mode				
SEC_RUN	N/A	01	Clocked	Clocked	Secondary – Timer1 Oscillator				
RC_RUN	N/A	11	Clocked	Clocked	Internal Oscillator				
PRI_IDLE	1	10	Off	Clocked	Primary – HS, EC, HSPLL, ECPLL				
SEC_IDLE	1	01	Off	Clocked	Secondary – Timer1 Oscillator				
RC_IDLE	1	11	Off	Clocked	Internal Oscillator				

TABLE 3-1: POWER-M	MANAGED MODES
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Note 1: IDLEN reflects its value when the SLEEP instruction is executed.

3.1.3 CLOCK TRANSITIONS AND STATUS INDICATORS

The length of the transition between clock sources is the sum of two cycles of the old clock source and three to four cycles of the new clock source. This formula assumes that the new clock source is stable.

Two bits indicate the current clock source and its status: OSTS (OSCCON<3>) and T1RUN (T1CON<6>). In general, only one of these bits will be set while in a given power-managed mode. When the OSTS bit is set, the primary clock is providing the device clock. When the T1RUN bit is set, the Timer1 oscillator is providing the clock. If neither of these bits is set, INTRC is clocking the device.

Note: Executing a SLEEP instruction does not necessarily place the device into Sleep mode. It acts as the trigger to place the controller into either the Sleep mode, or one of the Idle modes, depending on the setting of the IDLEN bit.

3.1.4 MULTIPLE SLEEP COMMANDS

The power-managed mode that is invoked with the SLEEP instruction is determined by the setting of the IDLEN bit at the time the instruction is executed. If another SLEEP instruction is executed, the device will enter the power-managed mode specified by IDLEN at that time. If IDLEN has changed, the device will enter the new power-managed mode specified by the new setting.

3.2 Run Modes

In the Run modes, clocks to both the core and peripherals are active. The difference between these modes is the clock source.

3.2.1 PRI_RUN MODE

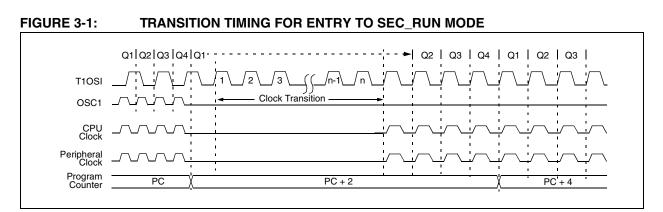
The PRI_RUN mode is the normal, full power execution mode of the microcontroller. This is also the default mode upon a device Reset unless Two-Speed Start-up is enabled (see Section 22.4 "Two-Speed Start-up" for details). In this mode, the OSTS bit is set (see Section 2.2 "Control Registers").

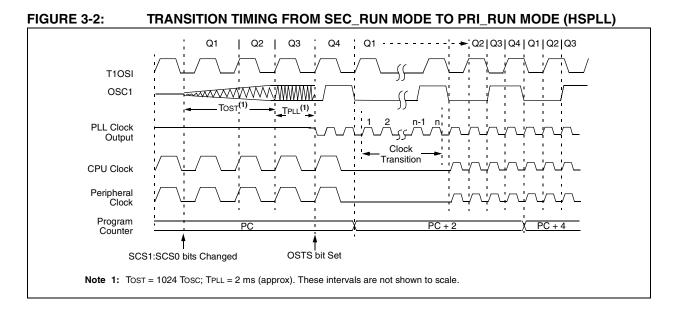
3.2.2 SEC_RUN MODE

The SEC_RUN mode is the compatible mode to the "clock switching" feature offered in other PIC18 devices. In this mode, the CPU and peripherals are clocked from the Timer1 oscillator. This gives users the option of lower power consumption while still using a high-accuracy clock source.

SEC_RUN mode is entered by setting the SCS1:SCS0 bits to '01'. The device clock source is switched to the Timer1 oscillator (see Figure 3-1), the primary oscillator is shut down, the T1RUN bit (T1CON<6>) is set and the OSTS bit is cleared.

Note: The Timer1 oscillator should already be running prior to entering SEC_RUN mode. If the T1OSCEN bit is not set when the SCS1:SCS0 bits are set to '01', entry to SEC_RUN mode will not occur. If the Timer1 oscillator is enabled, but not yet running, device clocks will be delayed until the oscillator has started. In such situations, initial oscillator operation is far from stable and unpredictable operation may result. On transitions from SEC_RUN mode to PRI_RUN mode, the peripherals and CPU continue to be clocked from the Timer1 oscillator while the primary clock is started. When the primary clock becomes ready, a clock switch back to the primary clock occurs (see Figure 3-2). When the clock switch is complete, the T1RUN bit is cleared, the OSTS bit is set and the primary clock is providing the clock. The IDLEN and SCS bits are not affected by the wake-up; the Timer1 oscillator continues to run.





3.2.3 RC_RUN MODE

In RC_RUN mode, the CPU and peripherals are clocked from the internal oscillator; the primary clock is shut down. This mode provides the best power conservation of all the Run modes while still executing code. It works well for user applications which are not highly timing sensitive or do not require high-speed clocks at all times.

This mode is entered by setting SCS bits to '11'. When the clock source is switched to the INTRC (see Figure 3-3), the primary oscillator is shut down and the OSTS bit is cleared. On transitions from RC_RUN mode to PRI_RUN mode, the device continues to be clocked from the INTRC while the primary clock is started. When the primary clock becomes ready, a clock switch to the primary clock occurs (see Figure 3-4). When the clock switch is complete, the OSTS bit is set and the primary clock is providing the device clock. The IDLEN and SCS bits are not affected by the switch. The INTRC source will continue to run if either the WDT or the Fail-Safe Clock Monitor is enabled.

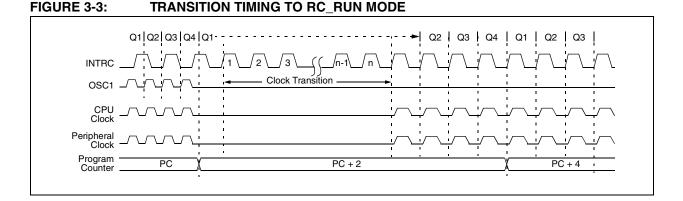
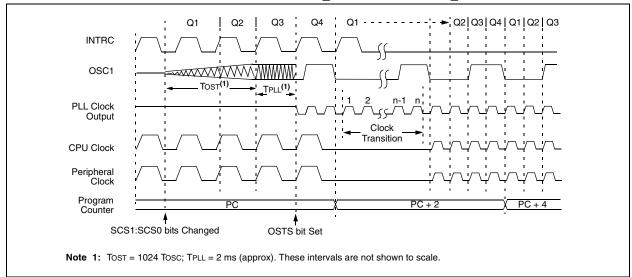


FIGURE 3-4: TRANSITION TIMING FROM RC RUN MODE TO PRI RUN MODE



3.3 Sleep Mode

The power-managed Sleep mode is identical to the legacy Sleep mode offered in all other PIC devices. It is entered by clearing the IDLEN bit (the default state on device Reset) and executing the SLEEP instruction. This shuts down the selected oscillator (Figure 3-5). All clock source status bits are cleared.

Entering the Sleep mode from any other mode does not require a clock switch. This is because no clocks are needed once the controller has entered Sleep. If the WDT is selected, the INTRC source will continue to operate. If the Timer1 oscillator is enabled, it will also continue to run.

When a wake event occurs in Sleep mode (by interrupt, Reset or WDT time-out), the device will not be clocked until the clock source selected by the SCS1:SCS0 bits becomes ready (see Figure 3-6), or it will be clocked from the internal oscillator if either the Two-Speed Start-up or the Fail-Safe Clock Monitor are enabled (see **Section 22.0 "Special Features of the CPU"**). In either case, the OSTS bit is set when the primary clock is providing the device clocks. The IDLEN and SCS bits are not affected by the wake-up.

3.4 Idle Modes

The Idle modes allow the controller's CPU to be selectively shut down while the peripherals continue to operate. Selecting a particular Idle mode allows users to further manage power consumption.

If the IDLEN bit is set to a '1' when a SLEEP instruction is executed, the peripherals will be clocked from the clock source selected using the SCS1:SCS0 bits; however, the CPU will not be clocked. The clock source status bits are not affected. Setting IDLEN and executing a SLEEP instruction provides a quick method of switching from a given Run mode to its corresponding Idle mode.

If the WDT is selected, the INTRC source will continue to operate. If the Timer1 oscillator is enabled, it will also continue to run.

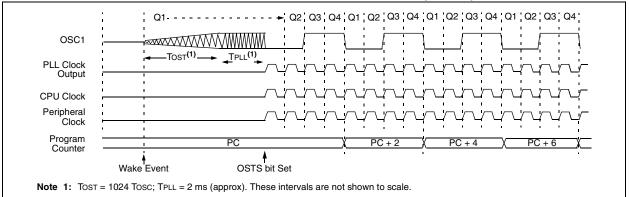
Since the CPU is not executing instructions, the only exits from any of the Idle modes are by interrupt, WDT time-out or a Reset. When a wake event occurs, CPU execution is delayed by an interval of TCSD (parameter 38, Table 25-11) while it becomes ready to execute code. When the CPU begins executing code, it resumes with the same clock source for the current Idle mode. For example, when waking from RC_IDLE mode, the internal oscillator block will clock the CPU and peripherals (in other words, RC_RUN mode). The IDLEN and SCS bits are not affected by the wake-up.

While in any Idle mode or the Sleep mode, a WDT time-out will result in a WDT wake-up to the Run mode currently specified by the SCS1:SCS0 bits.

FIGURE 3-5: TRANSITION TIMING FOR ENTRY TO SLEEP MODE

Q1 Q2 Q3 Q4	Q1		;		:			;	; ;	
			1	1	1 1		1	1		
			1	1	1		I	I		
			1	1	1 1		, ,	1 1		
Peripheral			1	1	1 1	1	1 1	1 1		
			1	1	1		1	1		
Sleep/			1	1	ı	ı	1 1	1 1	· ·	
Brogram	1		1	1	1		1	1		
Program Counter PC		PC + 2								





3.4.1 PRI_IDLE MODE

This mode is unique among the three low-power Idle modes, in that it does not disable the primary device clock. For timing sensitive applications, this allows for the fastest resumption of device operation with its more accurate primary clock source, since the clock source does not have to "warm up" or transition from another oscillator.

PRI_IDLE mode is entered from PRI_RUN mode by setting the IDLEN bit and executing a SLEEP instruction. If the device is in another Run mode, set IDLEN first, then set the SCS bits to '10' and execute SLEEP. Although the CPU is disabled, the peripherals continue to be clocked from the primary clock source specified by the FOSC1:FOSC0 Configuration bits. The OSTS bit remains set (see Figure 3-7).

When a wake event occurs, the CPU is clocked from the primary clock source. A delay of interval TCSD is required between the wake event and when code execution starts. This is required to allow the CPU to become ready to execute instructions. After the wake-up, the OSTS bit remains set. The IDLEN and SCS bits are not affected by the wake-up (see Figure 3-8).

3.4.2 SEC_IDLE MODE

In SEC_IDLE mode, the CPU is disabled but the peripherals continue to be clocked from the Timer1 oscillator. This mode is entered from SEC_RUN by setting the IDLEN bit and executing a SLEEP instruction. If the device is in another Run mode, set IDLEN first, then set SCS1:SCS0 to '01' and execute SLEEP. When the clock source is switched to the Timer1 oscillator, the primary oscillator is shut down, the OSTS bit is cleared and the T1RUN bit is set.

When a wake event occurs, the peripherals continue to be clocked from the Timer1 oscillator. After an interval of TCSD following the wake event, the CPU begins executing code being clocked by the Timer1 oscillator. The IDLEN and SCS bits are not affected by the wake-up; the Timer1 oscillator continues to run (see Figure 3-8).

Note: The Timer1 oscillator should already be running prior to entering SEC_IDLE mode. If the T1OSCEN bit is not set when the SLEEP instruction is executed, the SLEEP instruction will be ignored and entry to SEC_IDLE mode will not occur. If the Timer1 oscillator is enabled, but not yet running, peripheral clocks will be delayed until the oscillator has started. In such situations, initial oscillator operation is far from stable and unpredictable operation may result.

FIGURE 3-7: TRANSITION TIMING FOR ENTRY TO IDLE MODE Q1 Q2 Q3 Q4 Q1 OSC1 Q2 Q3 Q4 Q1 Q1

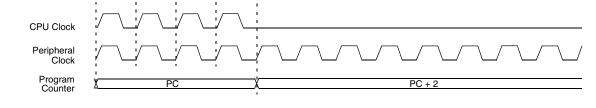
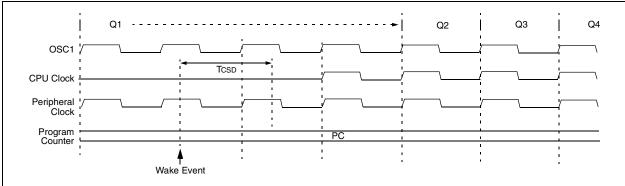


FIGURE 3-8: TRANSITION TIMING FOR WAKE FROM IDLE TO RUN MODE



3.4.3 RC_IDLE MODE

In RC_IDLE mode, the CPU is disabled but the peripherals continue to be clocked from the internal oscillator. This mode allows for controllable power conservation during Idle periods.

From RC_RUN, this mode is entered by setting the IDLEN bit and executing a SLEEP instruction. If the device is in another Run mode, first set IDLEN, then clear the SCS bits and execute SLEEP. When the clock source is switched to the INTRC, the primary oscillator is shut down and the OSTS bit is cleared.

When a wake event occurs, the peripherals continue to be clocked from the INTOSC. After a delay of TCSD following the wake event, the CPU begins executing code being clocked by the INTOSC. The IDLEN and SCS bits are not affected by the wake-up. The INTOSC source will continue to run if either the WDT or the Fail-Safe Clock Monitor is enabled.

3.5 Exiting Idle and Sleep Modes

An exit from Sleep mode, or any of the Idle modes, is triggered by an interrupt, a Reset or a WDT time-out. This section discusses the triggers that cause exits from power-managed modes. The clocking subsystem actions are discussed in each of the power-managed mode sections (see Section 3.2 "Run Modes", Section 3.3 "Sleep Mode" and Section 3.4 "Idle Modes").

3.5.1 EXIT BY INTERRUPT

Any of the available interrupt sources can cause the device to exit from an Idle mode, or the Sleep mode, to a Run mode. To enable this functionality, an interrupt source must be enabled by setting its enable bit in one of the INTCON or PIE registers. The exit sequence is initiated when the corresponding interrupt flag bit is set.

On all exits from Idle or Sleep modes by interrupt, code execution branches to the interrupt vector if the GIE/GIEH bit (INTCON<7>) is set. Otherwise, code execution continues or resumes without branching (see Section 8.0 "Interrupts").

A fixed delay of interval TCSD following the wake event is required when leaving Sleep and Idle modes. This delay is required for the CPU to prepare for execution. Instruction execution resumes on the first clock cycle following this delay.

3.5.2 EXIT BY WDT TIME-OUT

A WDT time-out will cause different actions depending on which power-managed mode the device is in when the time-out occurs.

If the device is not executing code (all Idle modes and Sleep mode), the time-out will result in an exit from the power-managed mode (see Section 3.2 "Run Modes" and Section 3.3 "Sleep Mode"). If the device is executing code (all Run modes), the time-out will result in a WDT Reset (see Section 22.2 "Watchdog Timer (WDT)").

The Watchdog Timer and postscaler are cleared by one of the following events:

- executing a SLEEP or CLRWDT instruction
- the loss of a currently selected clock source (if the Fail-Safe Clock Monitor is enabled)

3.5.3 EXIT BY RESET

Exiting an Idle or Sleep mode by Reset automatically forces the device to run from the INTRC.

3.5.4 EXIT WITHOUT AN OSCILLATOR START-UP DELAY

Certain exits from power-managed modes do not invoke the OST at all. There are two cases:

- PRI_IDLE mode, where the primary clock source is not stopped; and
- the primary clock source is either the EC or ECPLL mode.

In these instances, the primary clock source either does not require an oscillator start-up delay, since it is already running (PRI_IDLE), or normally does not require an oscillator start-up delay (EC). However, a fixed delay of interval TCSD following the wake event is still required when leaving Sleep and Idle modes to allow the CPU to prepare for execution. Instruction execution resumes on the first clock cycle following this delay.

NOTES:

4.0 RESET

The PIC18F85J90 family of devices differentiate between various kinds of Reset:

- a) Power-on Reset (POR)
- b) MCLR Reset during normal operation
- c) MCLR Reset during power-managed modes
- d) Watchdog Timer (WDT) Reset (during execution)
- e) Brown-out Reset (BOR)
- f) RESET Instruction
- g) Stack Full Reset
- h) Stack Underflow Reset

This section discusses Resets generated by MCLR, POR and BOR and covers the operation of the various start-up timers. Stack Reset events are covered in Section 5.1.4.4 "Stack Full and Underflow Resets". WDT Resets are covered in Section 22.2 "Watchdog Timer (WDT)".

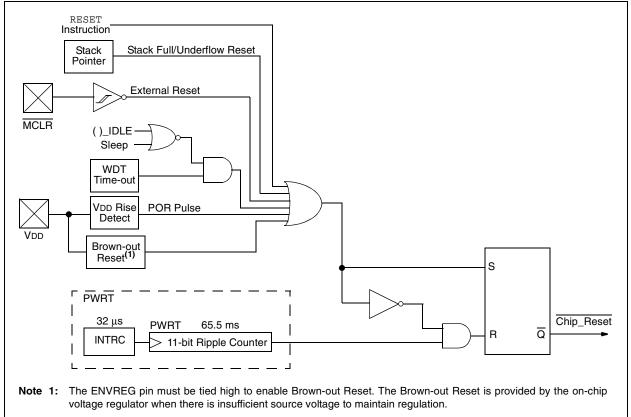
A simplified block diagram of the on-chip Reset circuit is shown in Figure 4-1.

4.1 RCON Register

Device Reset events are tracked through the RCON register (Register 4-1). The lower five bits of the register indicate that a specific Reset event has occurred. In most cases, these bits can only be set by the event and must be cleared by the application after the event. The state of these flag bits, taken together, can be read to indicate the type of Reset that just occurred. This is described in more detail in **Section 4.6 "Reset State of Registers"**.

The RCON register also has a control bit for setting interrupt priority (IPEN). Interrupt priority is discussed in **Section 8.0 "Interrupts"**.

FIGURE 4-1: SIMPLIFIED BLOCK DIAGRAM OF ON-CHIP RESET CIRCUIT



R/W-0	U-0	U-0	R/W-1	R-1	R-1	R/W-0	R/W-0
IPEN	_	_	RI	ТО	PD	POR	BOR
bit 7	·						bit 0
Legend:							
R = Readab -n = Value a		W = Writable '1' = Bit is set		0 = Unimpler	mented bit, rea	a as '0' x = Bit is unkr	2014/2
	ll FON				areu		IOWIT
bit 7	IPEN: Interrup	ot Priority Enal	ole bit				
	-	riority levels or					
	•	•	• •	IC16XXXX Cor	npatibility mode	e)	
bit 6-5	<u>Unimplemen</u>	ted: Read as '	0'				
bit 4		struction Flag b					
				ted (set by firm	• /	uat ha aat in aa	thucks offer a
		It Reset occurs		causing a dev	vice Reset (mi	ust be set in so	niware aller a
bit 3	TO: Watchdog	g Time-out Fla	g bit				
				or SLEEP instru	uction		
		me-out occurre					
bit 2		own Detection	•				
		ower-up or by t ecution of the					
bit 1		on Reset Statu					
	1 = A Power-	on Reset has	not occurred (s	set by firmware	only)		
						on Reset occur	s)
bit 0		out Reset Stat					
				(set by firmwar			
	0 = A Brown	-out Reset occ	currea (must be	e set in softwar	e alter a Brown	n-out Reset occi	urs)
			_				
	It is recommende Power-on Resets			er a Power-on F	Reset has beer	n detected, so th	at subsequent
	If the on-chip vo BOR" for more i		is disabled, B	OR remains '0	' at all times. S	See Section 4.4	.1 "Detecting
	Brown-out Reset				nd POR is '1' (a	assuming that \overline{P}	OR was set to

REGISTER 4-1: RCON: RESET CONTROL REGISTER

4.2 Master Clear (MCLR)

The MCLR pin provides a method for triggering a hard external Reset of the device. A Reset is generated by holding the pin low. PIC18 extended microcontroller devices have a noise filter in the MCLR Reset path which detects and ignores small pulses.

The $\overline{\text{MCLR}}$ pin is not driven low by any internal Resets, including the WDT.

4.3 Power-on Reset (POR)

A Power-on Reset condition is generated on-chip whenever VDD rises above a certain threshold. This allows the device to start in the initialized state when VDD is adequate for operation.

To take advantage of the POR circuitry, tie the $\overline{\text{MCLR}}$ pin through a resistor (1 k Ω to 10 k Ω) to VDD. This will eliminate external RC components usually needed to create a Power-on Reset delay. A minimum rise rate for VDD is specified (parameter D004). For a slow rise time, see Figure 4-2.

When the device starts normal operation (i.e., exits the Reset condition), device operating parameters (voltage, frequency, temperature, etc.) must be met to ensure operation. If these conditions are not met, the device must be held in Reset until the operating conditions are met.

Power-on Reset events are captured by the \overrightarrow{POR} bit (RCON<1>). The state of the bit is set to '0' whenever a Power-on Reset occurs; it does not change for any other Reset event. \overrightarrow{POR} is not reset to '1' by any hardware event. To capture multiple events, the user manually resets the bit to '1' in software following any Power-on Reset.

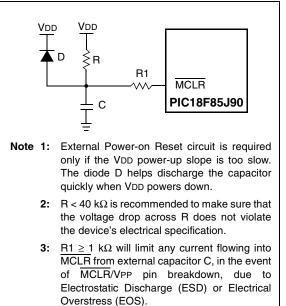
4.4 Brown-out Reset (BOR)

The PIC18F85J90 family of devices incorporates a simple BOR function when the internal regulator is enabled (ENVREG pin is tied to VDD). The voltage regulator will trigger a Brown-out Reset when output of the regulator to the device core approaches the voltage at which the device is unable to run at full speed. The BOR circuit also keeps the device in Reset as VDD rises, until the regulator's output level is sufficient for full-speed operation.

Once a BOR has occurred, the Power-up Timer will keep the chip in Reset for TPWRT (parameter 33). If VDD drops below the threshold for full-speed operation while the Power-up Timer is running, the chip will go back into a Brown-out Reset and the Power-up Timer will be initialized. Once VDD rises to the point where regulator output is sufficient, the Power-up Timer will execute the additional time delay.

FIGURE 4-2:

EXTERNAL POWER-ON RESET CIRCUIT (FOR SLOW VDD POWER-UP)



4.4.1 DETECTING BOR

The BOR bit always resets to '0' on any Brown-out Reset or Power-on Reset event. This makes it difficult to determine if a Brown-out Reset event has occurred just by reading the state of BOR alone. A more reliable method is to simultaneously check the state of both POR and BOR. This assumes that the POR bit is reset to '1' in software immediately after any Power-on Reset event. If BOR is '0' while POR is '1', it can be reliably assumed that a Brown-out Reset event has occurred.

If the voltage regulator is disabled, Brown-out Reset functionality is disabled. In this case, the $\overline{\text{BOR}}$ bit cannot be used to determine a Brown-out Reset event. The $\overline{\text{BOR}}$ bit is still cleared by a Power-on Reset event.

4.5 **Power-up Timer (PWRT)**

PIC18F85J90 family devices incorporate an on-chip Power-up Timer (PWRT) to help regulate the Power-on Reset process. The PWRT is always enabled. The main function is to ensure that the device voltage is stable before code is executed.

The Power-up Timer (PWRT) of the PIC18F85J90 family devices is an 11-bit counter which uses the INTRC source as the clock input. This yields an approximate time interval of 2048 x 32 μ s = 65.6 ms. While the PWRT is counting, the device is held in Reset.

The power-up time delay depends on the INTRC clock and will vary from chip-to-chip due to temperature and process variation. See DC parameter 33 for details.

4.5.1 TIME-OUT SEQUENCE

If enabled, the PWRT time-out is invoked after the POR pulse has cleared. The total time-out will vary based on the status of the PWRT. Figure 4-3, Figure 4-4, Figure 4-5 and Figure 4-6 all depict time-out sequences on power-up with the Power-up Timer enabled.

Since the time-outs occur from the POR pulse, if $\overline{\text{MCLR}}$ is kept low long enough, the PWRT will expire. Bringing $\overline{\text{MCLR}}$ high will begin execution immediately (Figure 4-5). This is useful for testing purposes, or to synchronize more than one PIC18FXXXX device operating in parallel.

FIGURE 4-3: TIME-OUT SEQUENCE ON POWER-UP (MCLR TIED TO VDD, VDD RISE < TPWRT)

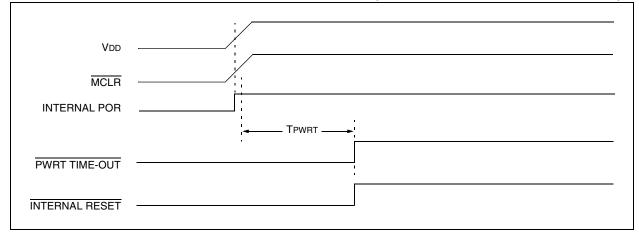
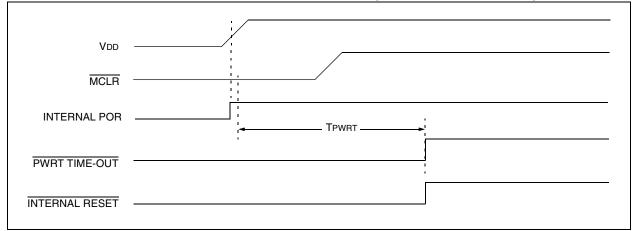


FIGURE 4-4: TIME-OUT SEQUENCE ON POWER-UP (MCLR NOT TIED TO VDD): CASE 1



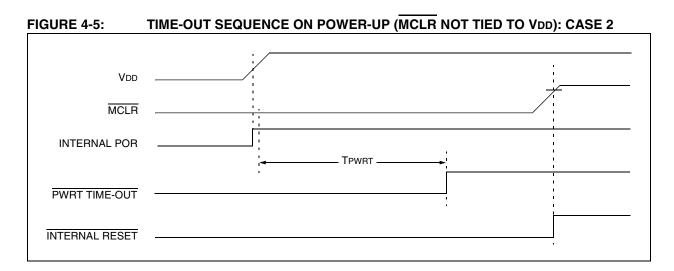
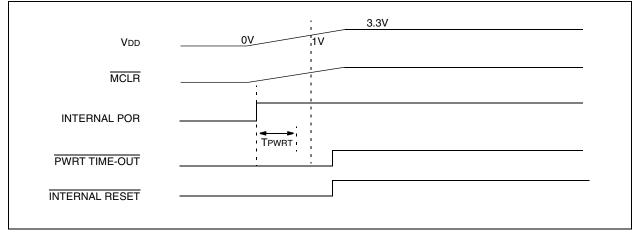


FIGURE 4-6: SLOW RISE TIME (MCLR TIED TO VDD, VDD RISE > TPWRT)



4.6 Reset State of Registers

Most registers are unaffected by a Reset. Their status is unknown on POR and unchanged by all other Resets. The other registers are forced to a "Reset state" depending on the type of Reset that occurred.

Most registers are not affected by a WDT wake-up, since this is viewed as the resumption of normal operation. Status bits from the RCON register, \overline{RI} , \overline{TO} , \overline{PD} , \overline{POR} and \overline{BOR} , are set or cleared differently in different Reset situations, as indicated in Table 4-1. These bits are used in software to determine the nature of the Reset.

Table 4-2 describes the Reset states for all of the Special Function Registers. These are categorized by Power-on and Brown-out Resets, Master Clear and WDT Resets and WDT wake-ups.

TABLE 4-1:	STATUS BITS, THEIR SIGNIFICANCE AND THE INITIALIZATION CONDITION FOR
	RCON REGISTER

O a maliki a m	Program		RC	STKPTR Register				
Condition	Counter ⁽¹⁾	RI	то	PD	POR	BOR	STKFUL	STKUNF
Power-on Reset	0000h	1	1	1	0	0	0	0
RESET Instruction	0000h	0	u	u	u	u	u	u
Brown-out Reset	0000h	1	1	1	u	0	u	u
MCLR during power-managed Run modes	0000h	u	1	u	u	u	u	u
MCLR during power-managed Idle modes and Sleep mode	0000h	u	1	0	u	u	u	u
WDT time-out during full power or power-managed Run modes	0000h	u	0	u	u	u	u	u
MCLR during full power execution	0000h	u	u	u	u	u	u	u
Stack Full Reset (STVREN = 1)	0000h	u	u	u	u	u	1	u
Stack Underflow Reset (STVREN = 1)	0000h	u	u	u	u	u	u	1
Stack Underflow Error (not an actual Reset, STVREN = 0)	0000h	u	u	u	u	u	u	1
WDT time-out during power-managed Idle or Sleep modes	PC + 2	u	0	0	u	u	u	u
Interrupt exit from power-managed modes	PC + 2	u	u	0	u	u	u	u

Legend: u = unchanged

Note 1: When the wake-up is due to an interrupt and the GIEH or GIEL bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

Register		e Devices	Power-on Reset, Brown-out Reset	MCLR Resets WDT Reset RESET Instruction Stack Resets	Wake-up via WDT or Interrupt	
TOSU	PIC18F6XJ90	PIC18F8XJ90	0 0000	0 0000	0 uuuu (1)	
TOSH	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu (1)	
TOSL	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu (1)	
STKPTR	PIC18F6XJ90	PIC18F8XJ90	uu-0 0000	00-0 0000	uu-u uuuu (1)	
PCLATU	PIC18F6XJ90	PIC18F8XJ90	0 0000	0 0000	u uuuu	
PCLATH	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu	
PCL	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	PC + 2 ⁽²⁾	
TBLPTRU	PIC18F6XJ90	PIC18F8XJ90	00 0000	00 0000	uu uuuu	
TBLPTRH	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu	
TBLPTRL	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu	
TABLAT	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu	
PRODH	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
PRODL	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
INTCON	PIC18F6XJ90	PIC18F8XJ90	0000 000x	0000 000u	uuuu uuuu (3)	
INTCON2	PIC18F6XJ90	PIC18F8XJ90	1111 1111	1111 1111	uuuu uuuu (3)	
INTCON3	PIC18F6XJ90	PIC18F8XJ90	1100 0000	1100 0000	uuuu uuuu ⁽³⁾	
INDF0	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A	
POSTINC0	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A	
POSTDEC0	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A	
PREINC0	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A	
PLUSW0	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A	
FSR0H	PIC18F6XJ90	PIC18F8XJ90	xxxx	uuuu	uuuu	
FSR0L	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	
WREG	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	
INDF1	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A	
POSTINC1	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A	
POSTDEC1	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A	
PREINC1	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A	
PLUSW1	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A	

TABLE 4-2:	INITIALIZATION CONDITIONS FOR ALL REGISTERS

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

Note 1: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

- 3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).
- 4: See Table 4-1 for Reset value for specific condition.

5: Bits 6 and 7 of PORTA, LATA and TRISA are enabled depending on the oscillator mode selected. When not enabled as PORTA pins, they are disabled and read as '0'.

TABLE 4-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)						
Register	Applicable Devices		Power-on Reset, Brown-out Reset	MCLR Resets WDT Reset RESET Instruction Stack Resets	Wake-up via WDT or Interrupt	
FSR1H	PIC18F6XJ90	PIC18F8XJ90	xxxx	uuuu	uuuu	
FSR1L	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	
BSR	PIC18F6XJ90	PIC18F8XJ90	0000	0000	uuuu	
INDF2	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A	
POSTINC2	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A	
POSTDEC2	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A	
PREINC2	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A	
PLUSW2	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A	
FSR2H	PIC18F6XJ90	PIC18F8XJ90	xxxx	uuuu	uuuu	
FSR2L	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	
STATUS	PIC18F6XJ90	PIC18F8XJ90	x xxxx	u uuuu	u uuuu	
TMR0H	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu	
TMR0L	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	
TOCON	PIC18F6XJ90	PIC18F8XJ90	1111 1111	1111 1111	uuuu uuuu	
OSCCON	PIC18F6XJ90	PIC18F8XJ90	0100 q000	0100 q000	uuuu quuu	
LCDREG	PIC18F6XJ90	PIC18F8XJ90	-011 1100	-011 1000	-uuu uuuu	
WDTCON	PIC18F6XJ90	PIC18F8XJ90	00	00	uu	
RCON ⁽⁴⁾	PIC18F6XJ90	PIC18F8XJ90	01 11q0	0q qquu	uu qquu	
TMR1H	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	
TMR1L	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	
T1CON	PIC18F6XJ90	PIC18F8XJ90	0000 0000	u0uu uuuu	uuuu uuuu	
TMR2	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu	
PR2	PIC18F6XJ90	PIC18F8XJ90	1111 1111	1111 1111	1111 1111	
T2CON	PIC18F6XJ90	PIC18F8XJ90	-000 0000	-000 0000	-uuu uuuu	
SSPBUF	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	
SSPADD	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu	
SSPSTAT	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu	
SSPCON1	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu	
SSPCON2	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu	

TABLE 4-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

 $\label{eq:logend:u} \begin{array}{ll} \mbox{Legend:} & u = \mbox{unchanged, $x = unknown, $- = unimplemented bit, read as `0', $q = value depends on condition.} \\ & Shaded cells indicate conditions do not apply for the designated device.} \end{array}$

Note 1: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

- **3:** One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).
- 4: See Table 4-1 for Reset value for specific condition.

5: Bits 6 and 7 of PORTA, LATA and TRISA are enabled depending on the oscillator mode selected. When not enabled as PORTA pins, they are disabled and read as '0'.

Register	Applicable Devices		Power-on Reset, Brown-out Reset	GISTERS (CONTINUE MCLR Resets WDT Reset RESET Instruction Stack Resets	Wake-up via WDT or Interrupt
ADRESH	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu
ADRESL	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu
ADCON0	PIC18F6XJ90	PIC18F8XJ90	0-00 0000	0-00 0000	u-uu uuuu
ADCON1	PIC18F6XJ90	PIC18F8XJ90	00 0000	00 0000	uu uuuu
ADCON2	PIC18F6XJ90	PIC18F8XJ90	0-00 0000	0-00 0000	u-uu uuuu
LCDDATA4	PIC18F6XJ90	PIC18F8XJ90	x	u	u
LCDDATA4	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	սսսս սսսս	uuuu uuuu
LCDDATA3	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu
LCDDATA2	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu
LCDDATA1	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu
LCDDATA0	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu
LCDSE5	PIC18F6XJ90	PIC18F8XJ90	0000 0000	uuuu uuuu	uuuu uuuu
LCDSE4	PIC18F6XJ90	PIC18F8XJ90	0	u	u
LCDSE4	PIC18F6XJ90	PIC18F8XJ90	0000 0000	uuuu uuuu	uuuu uuuu
LCDSE3	PIC18F6XJ90	PIC18F8XJ90	0000 0000	uuuu uuuu	uuuu uuuu
LCDSE2	PIC18F6XJ90	PIC18F8XJ90	0000 0000	uuuu uuuu	uuuu uuuu
LCDSE1	PIC18F6XJ90	PIC18F8XJ90	0000 0000	uuuu uuuu	uuuu uuuu
CVRCON	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
CMCON	PIC18F6XJ90	PIC18F8XJ90	0000 0111	0000 0111	uuuu uuuu
TMR3H	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu
TMR3L	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu
T3CON	PIC18F6XJ90	PIC18F8XJ90	0000 0000	uuuu uuuu	uuuu uuuu
SPBRG1	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
RCREG1	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
TXREG1	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
TXSTA1	PIC18F6XJ90	PIC18F8XJ90	0000 0010	0000 0010	uuuu uuuu
RCSTA1	PIC18F6XJ90	PIC18F8XJ90	0000 000x	0000 000x	uuuu uuuu
LCDPS	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
LCDSE0	PIC18F6XJ90	PIC18F8XJ90	0000 0000	uuuu uuuu	uuuu uuuu
LCDCON	PIC18F6XJ90	PIC18F8XJ90	000- 0000	000- 0000	uuu- uuuu
EECON2	PIC18F6XJ90	PIC18F8XJ90			
EECON1	PIC18F6XJ90	PIC18F8XJ90	0 x00-	0 u00-	0 u00-

TABLE 4-2:	INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

Note 1: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

- 2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).
- 3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).
- 4: See Table 4-1 for Reset value for specific condition.
- **5:** Bits 6 and 7 of PORTA, LATA and TRISA are enabled depending on the oscillator mode selected. When not enabled as PORTA pins, they are disabled and read as '0'.

TABLE 4-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)					ED)
Register	Applicabl	e Devices	Power-on Reset, Brown-out Reset	MCLR Resets WDT Reset RESET Instruction Stack Resets	Wake-up via WDT or Interrupt
IPR3	PIC18F6XJ90	PIC18F8XJ90	-111 -11-	-111 -11-	-uuu -uu-
PIR3	PIC18F6XJ90	PIC18F8XJ90	-000 -00-	-000 -00-	-uuu -00- (3)
PIE3	PIC18F6XJ90	PIC18F8XJ90	-000 -00-	-000 -00-	-uuu -00-
IPR2	PIC18F6XJ90	PIC18F8XJ90	11 111-	11 111-	uu uuu-
PIR2	PIC18F6XJ90	PIC18F8XJ90	00 000-	00 000-	uu uuu- ⁽³⁾
PIE2	PIC18F6XJ90	PIC18F8XJ90	00 000-	00 000-	uu uuu-
IPR1	PIC18F6XJ90	PIC18F8XJ90	-111 1-11	-111 1-11	-uuu u-uu
PIR1	PIC18F6XJ90	PIC18F8XJ90	-000 0-00	-000 0-00	-uuu u-uu ⁽³⁾
PIE1	PIC18F6XJ90	PIC18F8XJ90	-000 0-00	-000 0-00	-uuu u-uu
OSCTUNE	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
TRISJ	PIC18F6XJ90	PIC18F8XJ90	1111 1111	1111 1111	uuuu uuuu
TRISH	PIC18F6XJ90	PIC18F8XJ90	1111 1111	1111 1111	uuuu uuuu
TRISG	PIC18F6XJ90	PIC18F8XJ90	0001 1111	0001 1111	uuuu uuuu
TRISF	PIC18F6XJ90	PIC18F8XJ90	1111 111-	1111 111-	uuuu uuu-
TRISE	PIC18F6XJ90	PIC18F8XJ90	1111 1-11	1111 1-11	uuuu u-uu
TRISD	PIC18F6XJ90	PIC18F8XJ90	1111 1111	1111 1111	uuuu uuuu
TRISC	PIC18F6XJ90	PIC18F8XJ90	1111 1111	1111 1111	uuuu uuuu
TRISB	PIC18F6XJ90	PIC18F8XJ90	1111 1111	1111 1111	uuuu uuuu
TRISA ⁽⁵⁾	PIC18F6XJ90	PIC18F8XJ90	1111 1111 (5)	1111 1111 (5)	uuuu uuuu (5)
LATJ	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu
LATH	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu
LATG	PIC18F6XJ90	PIC18F8XJ90	00-x xxxx	00-u uuuu	uu-u uuuu
LATF	PIC18F6XJ90	PIC18F8XJ90	xxxx xxx-	uuuu uuu-	uuuu uuu-
LATE	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu
LATD	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu
LATC	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu
LATB	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu
LATA ⁽⁵⁾	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx(5)	uuuu uuuu (5)	uuuu uuuu (5)
PORTJ	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu
PORTH	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu
PORTG	PIC18F6XJ90	PIC18F8XJ90	000x xxxx	000u uuuu	000u uuuu
PORTF	PIC18F6XJ90	PIC18F8XJ90	xxxx xxx-	uuuu uuu-	uuuu uuu-

TABLE 4-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

Note 1: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

- 2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).
- 3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).
- 4: See Table 4-1 for Reset value for specific condition.
- 5: Bits 6 and 7 of PORTA, LATA and TRISA are enabled depending on the oscillator mode selected. When not enabled as PORTA pins, they are disabled and read as '0'.

Register	Applicable Devices		Power-on Reset, Brown-out Reset	MCLR Resets WDT Reset RESET Instruction Stack Resets	Wake-up via WDT or Interrupt	
PORTE	PIC18F6XJ90	PIC18F8XJ90	xxxx x-xx	uuuu u-uu	uuuu u-uu	
PORTD	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
PORTC	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	սսսս սսսս	uuuu uuuu	
PORTB	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
PORTA ⁽⁵⁾	PIC18F6XJ90	PIC18F8XJ90	xx0x 0000 (5)	uu0u 0000 (5)	uuuu uuuu (5)	
SPBRGH1	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu	
BAUDCON1	PIC18F6XJ90	PIC18F8XJ90	01-0 0-00	01-0 0-00	uu-u u-uu	
LCDDATA23	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA22	PIC18F6XJ90	PIC18F8XJ90	x	u	u	
LCDDATA22	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA21	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA20	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA19	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA18	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA17	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA16	PIC18F6XJ90	PIC18F8XJ90	x	u	u	
LCDDATA16	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA15	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA14	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA13	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	
LCDDATA12	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA11	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA10	PIC18F6XJ90	PIC18F8XJ90	x	u	u	
LCDDATA10	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA9	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA8	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	
LCDDATA7	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA6	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	
LCDDATA5	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
CCPR1H	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	
CCPR1L	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
CCP1CON	PIC18F6XJ90	PIC18F8XJ90	00 0000	00 0000	uu uuuu	
CCPR2H	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	

TABLE 4-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

 $\label{eq:logistical_logistical$

Note 1: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

- 2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).
- 3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).
- 4: See Table 4-1 for Reset value for specific condition.
- 5: Bits 6 and 7 of PORTA, LATA and TRISA are enabled depending on the oscillator mode selected. When not enabled as PORTA pins, they are disabled and read as '0'.

Register	Applicable Devices		Power-on Reset, Brown-out Reset	MCLR Resets WDT Reset RESET Instruction Stack Resets	Wake-up via WDT or Interrupt
CCPR2L	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu
CCP2CON	PIC18F6XJ90	PIC18F8XJ90	00 0000	00 0000	uu uuuu
SPBRG2	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
RCREG2	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
TXREG2	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
TXSTA2	PIC18F6XJ90	PIC18F8XJ90	0000 -010	0000 -010	uuuu -uuu
RCSTA2	PIC18F6XJ90	PIC18F8XJ90	0000 000x	0000 000x	uuuu uuuu

TABLE 4-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

Note 1: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

- 3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).
- 4: See Table 4-1 for Reset value for specific condition.
- **5:** Bits 6 and 7 of PORTA, LATA and TRISA are enabled depending on the oscillator mode selected. When not enabled as PORTA pins, they are disabled and read as '0'.

5.0 MEMORY ORGANIZATION

There are two types of memory in PIC18 Flash microcontroller devices:

- Program Memory
- Data RAM

As Harvard architecture devices, the data and program memories use separate busses; this allows for concurrent access of the two memory spaces.

Additional detailed information on the operation of the Flash program memory is provided in **Section 6.0 "Flash Program Memory"**.

5.1 Program Memory Organization

PIC18 microcontrollers implement a 21-bit program counter which is capable of addressing a 2-Mbyte program memory space. Accessing a location between the upper boundary of the physically implemented memory and the 2-Mbyte address will return all '0's (a NOP instruction).

The entire PIC18F85J90 family offers a range of on-chip Flash program memory sizes, from 8 Kbytes (up to 4,096 single-word instructions) to 32 Kbytes (32,768 single-word instructions). The program memory maps for individual family members are shown in Figure 5-1.

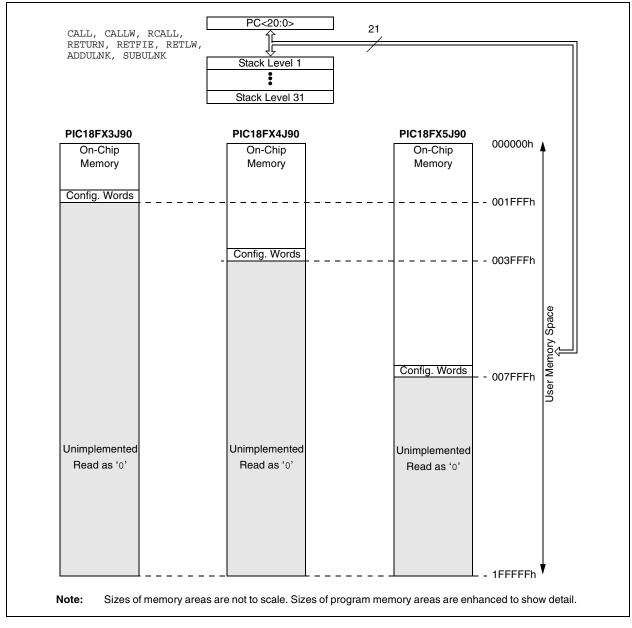


FIGURE 5-1: MEMORY MAPS FOR PIC18F85J90 FAMILY DEVICES

5.1.1 HARD MEMORY VECTORS

All PIC18 devices have a total of three hard-coded return vectors in their program memory space. The Reset vector address is the default value to which the program counter returns on all device Resets; it is located at 0000h.

PIC18 devices also have two interrupt vector addresses for the handling of high priority and low priority interrupts. The high priority interrupt vector is located at 0008h and the low priority interrupt vector is at 0018h. Their locations in relation to the program memory map are shown in Figure 5-2.

FIGURE 5-2: HARD VECTOR AND CONFIGURATION WORD LOCATIONS FOR PIC18F85J90 FAMILY FAMILY DEVICES

Reset Vector	0000h
High Priority Interrupt Vector	0008h
Low Priority Interrupt Vector	0018h
On-Chip Program Memory	
Flash Configuration Words	(Top of Memory-7) (Top of Memory)
Read 'o'	
	1FFFFFh
Legend: (Top of Memory) repression of on-chip program m Figure 5-1 for device- Shaded area represse memory. Areas are no	specific values). nts unimplemented

5.1.2 FLASH CONFIGURATION WORDS

Because PIC18F85J90 family devices do not have persistent configuration memory, the top four words of on-chip program memory are reserved for configuration information. On Reset, the configuration information is copied into the Configuration registers.

The Configuration Words are stored in their program memory location in numerical order, starting with the lower byte of CONFIG1 at the lowest address and ending with the upper byte of CONFIG4. For these devices, only Configuration Words, CONFIG1 through CONFIG3, are used; CONFIG4 is reserved. The actual addresses of the Flash Configuration Word for devices in the PIC18F85J90 family are shown in Table 5-1. Their location in the memory map is shown with the other memory vectors in Figure 5-2.

Additional details on the device Configuration Words are provided in **Section 22.1** "Configuration Bits".

TABLE 5-1:	FLASH CONFIGURATION
	WORD FOR PIC18F85J90
	FAMILY DEVICES

Device	Program Memory (Kbytes)	Configuration Word Addresses
PIC18F63J90	8	1FF8h to 1FFFh
PIC18F83J90	0	
PIC18F64J90	16	3FF8h to 3FFFh
PIC18F84J90	10	355011 10 35571
PIC18F65J90	32	7FF8h to 7FFFh
PIC18F85J90	52	

5.1.3 PROGRAM COUNTER

The Program Counter (PC) specifies the address of the instruction to fetch for execution. The PC is 21 bits wide and is contained in three separate 8-bit registers. The low byte, known as the PCL register, is both readable and writable. The high byte, or PCH register, contains the PC<15:8> bits; it is not directly readable or writable. Updates to the PCH register are performed through the PCLATH register. The upper byte is called PCU. This register contains the PC<20:16> bits; it is also not directly readable or writable. Updates to the PCH register. Updates to the PCU register are performed through the PCLATH register contains the PC<20:16> bits; it is also not directly readable or writable. Updates to the PCU register are performed through the PCLATU register.

The contents of PCLATH and PCLATU are transferred to the program counter by any operation that writes PCL. Similarly, the upper two bytes of the program counter are transferred to PCLATH and PCLATU by an operation that reads PCL. This is useful for computed offsets to the PC (see **Section 5.1.6.1 "Computed GOTO**").

The PC addresses bytes in the program memory. To prevent the PC from becoming misaligned with word instructions, the Least Significant bit of PCL is fixed to a value of '0'. The PC increments by 2 to address sequential instructions in the program memory.

The CALL, RCALL, GOTO and program branch instructions write to the program counter directly. For these instructions, the contents of PCLATH and PCLATU are not transferred to the program counter.

5.1.4 RETURN ADDRESS STACK

The return address stack allows any combination of up to 31 program calls and interrupts to occur. The PC is pushed onto the stack when a CALL or RCALL instruction is executed, or an interrupt is Acknowledged. The PC value is pulled off the stack on a RETURN, RETLW or a RETFIE instruction (and on ADDULNK and SUBULNK instructions if the extended instruction set is enabled). PCLATU and PCLATH are not affected by any of the RETURN or CALL instructions. The stack operates as a 31-word by 21-bit RAM and a 5-bit Stack Pointer, STKPTR. The stack space is not part of either program or data space. The Stack Pointer is readable and writable and the address on the top of the stack is readable and writable through the Top-of-Stack Special Function Registers. Data can also be pushed to, or popped from the stack, using these registers.

A CALL type instruction causes a push onto the stack. The Stack Pointer is first incremented and the location pointed to by the Stack Pointer is written with the contents of the PC (already pointing to the instruction following the CALL). A RETURN type instruction causes a pop from the stack. The contents of the location pointed to by the STKPTR are transferred to the PC and then the Stack Pointer is decremented.

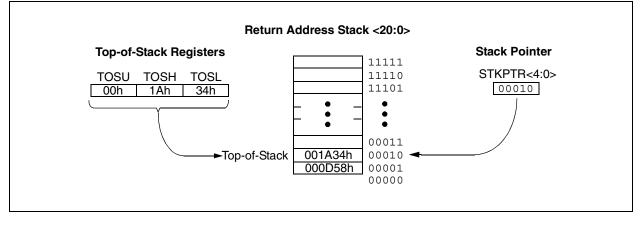
The Stack Pointer is initialized to '00000' after all Resets. There is no RAM associated with the location corresponding to a Stack Pointer value of '00000'; this is only a Reset value. Status bits indicate if the stack is full, has overflowed or has underflowed.

5.1.4.1 Top-of-Stack Access

Only the top of the return address stack (TOS) is readable and writable. A set of three registers, TOSU:TOSH:TOSL, holds the contents of the stack location pointed to by the STKPTR register (Figure 5-3). This allows users to implement a software stack if necessary. After a CALL, RCALL or interrupt (and ADDULNK and SUBULNK instructions if the extended instruction set is enabled), the software can read the pushed value by reading the TOSU:TOSH:TOSL registers. These values can be placed on a user-defined software stack. At return time, the software can return these values to TOSU:TOSH:TOSL and do a return.

The user must disable the global interrupt enable bits while accessing the stack to prevent inadvertent stack corruption.





5.1.4.2 Return Stack Pointer (STKPTR)

The STKPTR register (Register 5-1) contains the Stack Pointer value, the STKFUL (Stack Full) status bit and the STKUNF (Stack Underflow) status bits. The value of the Stack Pointer can be 0 through 31. The Stack Pointer increments before values are pushed onto the stack and decrements after values are popped off the stack. On Reset, the Stack Pointer value will be zero. The user may read and write the Stack Pointer value. This feature can be used by a Real-Time Operating System (RTOS) for return stack maintenance.

After the PC is pushed onto the stack 31 times (without popping any values off the stack), the STKFUL bit is set. The STKFUL bit is cleared by software or by a POR.

The action that takes place when the stack becomes full depends on the state of the STVREN (Stack Overflow Reset Enable) Configuration bit. (Refer to **Section 22.1 "Configuration Bits**" for a description of the device Configuration bits.) If STVREN is set (default), the 31st push will push the (PC + 2) value onto the stack, set the STKFUL bit and reset the device. The STKFUL bit will remain set and the Stack Pointer will be set to zero.

If STVREN is cleared, the STKFUL bit will be set on the 31st push and the Stack Pointer will increment to 31. Any additional pushes will not overwrite the 31st push and the STKPTR will remain at 31.

When the stack has been popped enough times to unload the stack, the next pop will return a value of zero to the PC and set the STKUNF bit, while the Stack Pointer remains at zero. The STKUNF bit will remain set until cleared by software or until a POR occurs.

Note:	Returning a value of zero to the PC on an underflow has the effect of vectoring the program to the Reset vector, where the stack conditions can be verified and appropriate actions can be taken. This is not the same as a Reset, as the contents
	of the SFRs are not affected.

5.1.4.3 PUSH and POP Instructions

Since the Top-of-Stack is readable and writable, the ability to push values onto the stack and pull values off the stack, without disturbing normal program execution, is a desirable feature. The PIC18 instruction set includes two instructions, PUSH and POP, that permit the TOS to be manipulated under software control. TOSU, TOSH and TOSL can be modified to place data or a return address on the stack.

The PUSH instruction places the current PC value onto the stack. This increments the Stack Pointer and loads the current PC value onto the stack.

The POP instruction discards the current TOS by decrementing the Stack Pointer. The previous value pushed onto the stack then becomes the TOS value.

R/C-0 R/C-0 U-0 **R/W-0 R/W-0** R/W-0 R/W-0 R/W-0 STKFUL⁽¹⁾ STKUNF⁽¹⁾ SP4 SP3 SP2 SP1 SP0 bit 7 bit 0

REGISTER 5-1: STKPTR: STACK POINTER REGISTER

Legend:	C = Clearable-only bit		
R = Readable bit	W = Writable bit	U = Unimplemented bit	t, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	STKFUL: Stack Full Flag bit ⁽¹⁾
	1 = Stack became full or overflowed
	0 = Stack has not become full or overflowed
bit 6	STKUNF: Stack Underflow Flag bit ⁽¹⁾
	1 = Stack underflow occurred
	0 = Stack underflow did not occur
bit 5	Unimplemented: Read as '0'
bit 4-0	SP4:SP0: Stack Pointer Location bits

Note 1: Bit 7 and bit 6 are cleared by user software or by a POR.

5.1.4.4 Stack Full and Underflow Resets

Device Resets on stack overflow and stack underflow conditions are enabled by setting the STVREN bit in Configuration Register 1L. When STVREN is set, a full or underflow condition will set the appropriate STKFUL or STKUNF bit and then cause a device Reset. When STVREN is cleared, a full or underflow condition will set the appropriate STKFUL or STKUNF bit, but not cause a device Reset. The STKFUL or STKUNF bits are cleared by the user software or a Power-on Reset.

5.1.5 FAST REGISTER STACK

A Fast Register Stack is provided for the STATUS, WREG and BSR registers to provide a "fast return" option for interrupts. This stack is only one level deep and is neither readable nor writable. It is loaded with the current value of the corresponding register when the processor vectors for an interrupt. All interrupt sources will push values into the Stack registers. The values in the registers are then loaded back into the working registers if the RETFIE, FAST instruction is used to return from the interrupt.

If both low and high priority interrupts are enabled, the Stack registers cannot be used reliably to return from low priority interrupts. If a high priority interrupt occurs while servicing a low priority interrupt, the Stack register values stored by the low priority interrupt will be overwritten. In these cases, users must save the key registers in software during a low priority interrupt.

If interrupt priority is not used, all interrupts may use the Fast Register Stack for returns from interrupt. If no interrupts are used, the Fast Register Stack can be used to restore the STATUS, WREG and BSR registers at the end of a subroutine call. To use the Fast Register Stack for a subroutine call, a CALL label, FAST instruction must be executed to save the STATUS, WREG and BSR registers to the Fast Register Stack. A RETURN, FAST instruction is then executed to restore these registers from the Fast Register Stack.

Example 5-1 shows a source code example that uses the Fast Register Stack during a subroutine call and return.

EXAMPLE 5-1: FAST REGISTER STACK CODE EXAMPLE

CALL SUB1, FAST • •	;STATUS, WREG, BSR ;SAVED IN FAST REGISTER ;STACK
SUB1 •	
RETURN FAST	;RESTORE VALUES SAVED ;IN FAST REGISTER STACK

5.1.6 LOOK-UP TABLES IN PROGRAM MEMORY

There may be programming situations that require the creation of data structures, or look-up tables, in program memory. For PIC18 devices, look-up tables can be implemented in two ways:

- Computed GOTO
- Table Reads

5.1.6.1 Computed GOTO

A computed GOTO is accomplished by adding an offset to the program counter. An example is shown in Example 5-2.

A look-up table can be formed with an ADDWF PCL instruction and a group of RETLW nn instructions. The W register is loaded with an offset into the table before executing a call to that table. The first instruction of the called routine is the ADDWF PCL instruction. The next instruction executed will be one of the RETLW nn instructions that returns the value 'nn' to the calling function.

The offset value (in WREG) specifies the number of bytes that the program counter should advance and should be multiples of 2 (LSb = 0).

In this method, only one data byte may be stored in each instruction location and room on the return address stack is required.

EXAMPLE 5-2: COMPUTED GOTO USING AN OFFSET VALUE

	MOVF	OFFSET, W
	CALL	TABLE
ORG	nn00h	
TABLE	ADDWF	PCL
	RETLW	nnh
	RETLW	nnh
	RETLW	nnh

5.1.6.2 Table Reads

A better method of storing data in program memory allows two bytes of data to be stored in each instruction location.

Look-up table data may be stored two bytes per program word while programming. The Table Pointer (TBLPTR) specifies the byte address and the Table Latch (TABLAT) contains the data that is read from the program memory. Data is transferred from program memory one byte at a time.

Table read operation is discussed further in **Section 6.1 "Table Reads and Table Writes"**.

5.2 PIC18 Instruction Cycle

5.2.1 CLOCKING SCHEME

The microcontroller clock input, whether from an internal or external source, is internally divided by four to generate four non-overlapping quadrature clocks (Q1, Q2, Q3 and Q4). Internally, the program counter is incremented on every Q1; the instruction is fetched from the program memory and latched into the Instruction Register (IR) during Q4. The instruction is decoded and executed during the following Q1 through Q4. The clocks and instruction execution flow are shown in Figure 5-4.

5.2.2 INSTRUCTION FLOW/PIPELINING

An "Instruction Cycle" consists of four Q cycles, Q1 through Q4. The instruction fetch and execute are pipelined in such a manner that a fetch takes one instruction cycle, while the decode and execute take another instruction cycle. However, due to the pipelining, each instruction effectively executes in one cycle. If an instruction causes the program counter to change (e.g., GOTO), then two cycles are required to complete the instruction (Example 5-3).

A fetch cycle begins with the Program Counter (PC) incrementing in Q1.

In the execution cycle, the fetched instruction is latched into the Instruction Register (IR) in cycle Q1. This instruction is then decoded and executed during the Q2, Q3 and Q4 cycles. Data memory is read during Q2 (operand read) and written during Q4 (destination write).

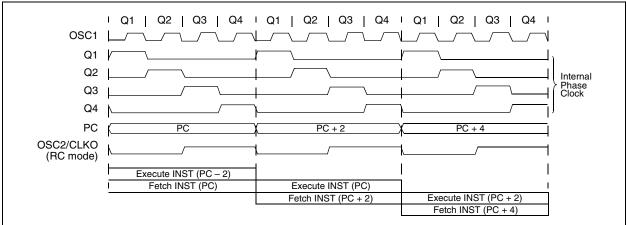
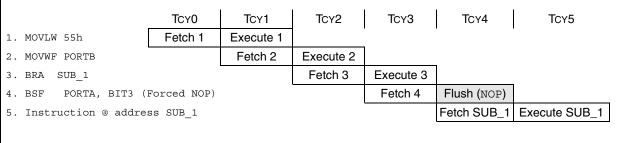


FIGURE 5-4: CLOCK/INSTRUCTION CYCLE

EXAMPLE 5-3: INSTRUCTION PIPELINE FLOW



All instructions are single cycle, except for any program branches. These take two cycles since the fetch instruction is "flushed" from the pipeline while the new instruction is being fetched and then executed.

5.2.3 INSTRUCTIONS IN PROGRAM MEMORY

The program memory is addressed in bytes. Instructions are stored as two bytes or four bytes in program memory. The Least Significant Byte of an instruction word is always stored in a program memory location with an even address (LSB = 0). To maintain alignment with instruction boundaries, the PC increments in steps of 2 and the LSB will always read '0' (see Section 5.1.3 "Program Counter").

Figure 5-5 shows an example of how instruction words are stored in the program memory.

The CALL and GOTO instructions have the absolute program memory address embedded into the instruction. Since instructions are always stored on word boundaries, the data contained in the instruction is a word address. The word address is written to PC<20:1> which accesses the desired byte address in program memory. Instruction #2 in Figure 5-5 shows how the instruction, GOTO 0006h, is encoded in the program memory. Program branch instructions, which encode a relative address offset, operate in the same manner. The offset value stored in a branch instruction represents the number of single-word instructions that the PC will be offset by. **Section 23.0 "Instruction Set Summary"** provides further details of the instruction set.

FIGURE 5-5:	INSTRUCTIONS IN PROGRAM MEMORY
-------------	--------------------------------

			LSB = 1	LSB = 0	Word Address \downarrow
	Program N	lemory			000000h
	Byte Locat	ions \rightarrow			000002h
					000004h
					000006h
Instruction 1:	MOVLW	055h	0Fh	55h	000008h
Instruction 2:	GOTO	0006h	EFh	03h	00000Ah
			F0h	00h	00000Ch
Instruction 3:	MOVFF	123h, 456h	C1h	23h	00000Eh
			F4h	56h	000010h
					000012h
					000014h

5.2.4 TWO-WORD INSTRUCTIONS

The standard PIC18 instruction set has four two-word instructions: CALL, MOVFF, GOTO and LSFR. In all cases, the second word of the instructions always has '1111' as its four Most Significant bits; the other 12 bits are literal data, usually a data memory address.

The use of '1111' in the 4 MSbs of an instruction specifies a special form of NOP. If the instruction is executed in proper sequence – immediately after the first word – the data in the second word is accessed

and used by the instruction sequence. If the first word is skipped for some reason and the second word is executed by itself, a NOP is executed instead. This is necessary for cases when the two-word instruction is preceded by a conditional instruction that changes the PC. Example 5-4 shows how this works.

Note:	See Section 5.5 "Program Memory and
	the Extended Instruction Set" for
	information on two-word instructions in the
	extended instruction set.

EXAMPLE 5-4: TWO-WORD INSTRUCTIONS

CASE 1:		
Object Code	Source Code	
0110 0110 0000 0000	TSTFSZ REG1	; is RAM location 0?
1100 0001 0010 0011	MOVFF REG1, REG2	; No, skip this word
1111 0100 0101 0110		; Execute this word as a NOP
0010 0100 0000 0000	ADDWF REG3	; continue code
CASE 2:		
Object Code	Source Code	
0110 0110 0000 0000	TSTFSZ REG1	; is RAM location 0?
1100 0001 0010 0011	MOVFF REG1, REG2	; Yes, execute this word
1111 0100 0101 0110		; 2nd word of instruction
0010 0100 0000 0000	ADDWF REG3	; continue code

5.3 Data Memory Organization

Note:	The operation of some aspects of data
	memory are changed when the PIC18
	extended instruction set is enabled. See
	Section 5.6 "Data Memory and the
	Extended Instruction Set" for more
	information.

The data memory in PIC18 devices is implemented as static RAM. Each register in the data memory has a 12-bit address, allowing up to 4096 bytes of data memory. The memory space is divided into as many as 16 banks that contain 256 bytes each. The PIC18FX3J90/X4J90 devices, with up to 16 Kbytes of program memory, implement 4 complete banks for a total of 1024 bytes. PIC18FX5J90 devices, with 32 Kbytes of program memory, implement 8 complete banks for a total of 2048 bytes. Figure 5-6 and Figure 5-7 show the data memory organization for the devices.

The data memory contains Special Function Registers (SFRs) and General Purpose Registers (GPRs). The SFRs are used for control and status of the controller and peripheral functions, while GPRs are used for data storage and scratchpad operations in the user's application. Any read of an unimplemented location will read as '0's.

The instruction set and architecture allow operations across all banks. The entire data memory may be accessed by Direct, Indirect or Indexed Addressing modes. Addressing modes are discussed later in this section.

To ensure that commonly used registers (select SFRs and select GPRs) can be accessed in a single cycle, PIC18 devices implement an Access Bank. This is a 256-byte memory space that provides fast access to select SFRs and the lower portion of GPR Bank 0 without using the BSR. **Section 5.3.2 "Access Bank**" provides a detailed description of the Access RAM.

5.3.1 BANK SELECT REGISTER

Large areas of data memory require an efficient addressing scheme to make rapid access to any address possible. Ideally, this means that an entire address does not need to be provided for each read or write operation. For PIC18 devices, this is accomplished with a RAM banking scheme. This divides the memory space into 16 contiguous banks of 256 bytes. Depending on the instruction, each location can be addressed directly by its full 12-bit address, or an 8-bit low-order address and a 4-bit Bank Pointer.

Most instructions in the PIC18 instruction set make use of the Bank Pointer, known as the Bank Select Register (BSR). This SFR holds the 4 Most Significant bits of a location's address; the instruction itself includes the 8 Least Significant bits. Only the four lower bits of the BSR are implemented (BSR3:BSR0). The upper four bits are unused; they will always read '0' and cannot be written to. The BSR can be loaded directly by using the MOVLB instruction.

The value of the BSR indicates the bank in data memory. The 8 bits in the instruction show the location in the bank and can be thought of as an offset from the bank's lower boundary. The relationship between the BSR's value and the bank division in data memory is shown in Figure 5-8.

Since up to 16 registers may share the same low-order address, the user must always be careful to ensure that the proper bank is selected before performing a data read or write. For example, writing what should be program data to an 8-bit address of F9h while the BSR is 0Fh, will end up resetting the program counter.

While any bank can be selected, only those banks that are actually implemented can be read or written to. Writes to unimplemented banks are ignored, while reads from unimplemented banks will return '0's. Even so, the STATUS register will still be affected as if the operation was successful. The data memory map in Figure 5-6 indicates which banks are implemented.

In the core PIC18 instruction set, only the MOVFF instruction fully specifies the 12-bit address of the source and target registers. This instruction ignores the BSR completely when it executes. All other instructions include only the low-order address as an operand and must use either the BSR or the Access Bank to locate their target registers.

FIGURE 5-6: DATA MEMORY MAP FOR PIC18FX3J90/X4J90 DEVICES

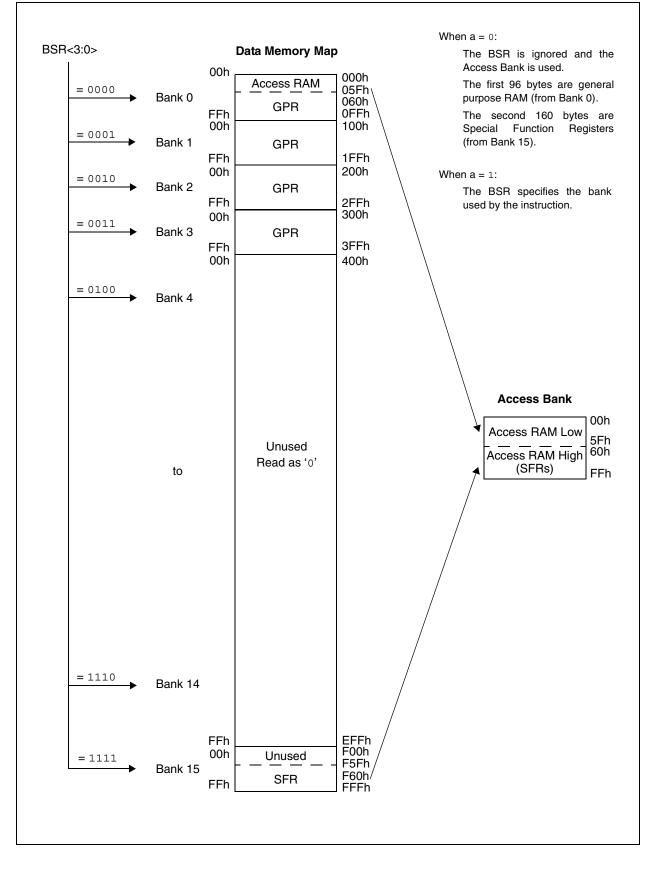
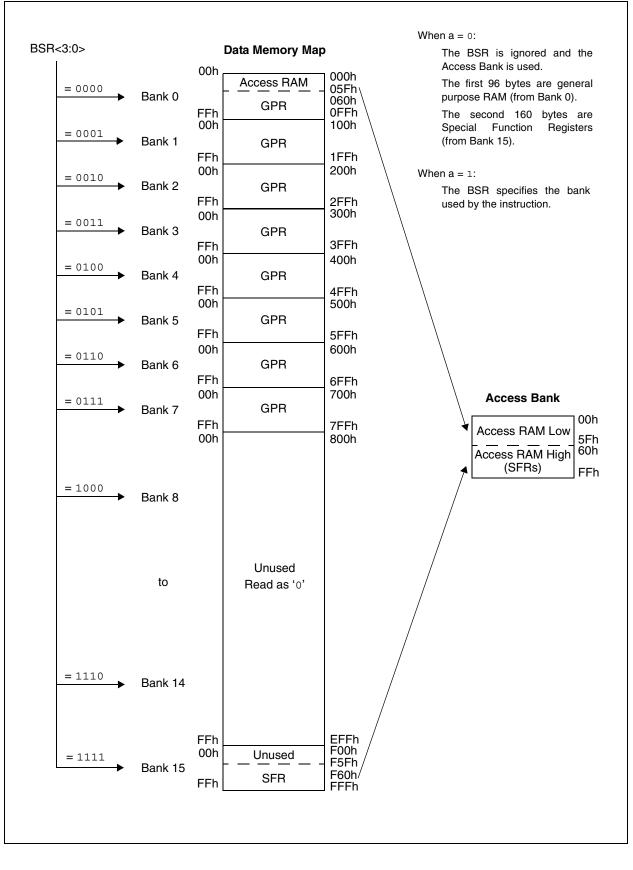
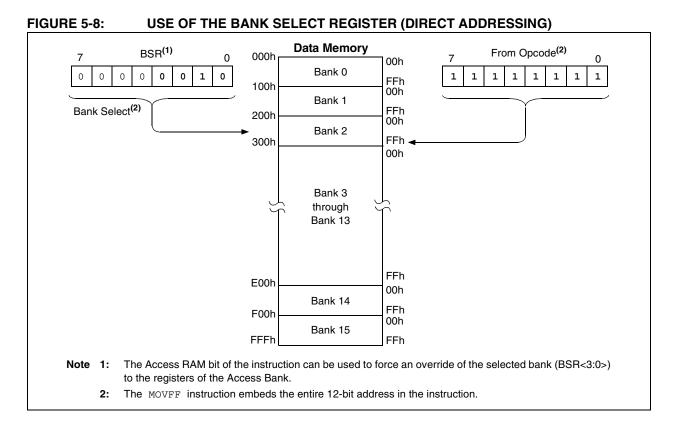


FIGURE 5-7: DATA MEMORY MAP FOR PIC18FX5J90 DEVICES





5.3.2 ACCESS BANK

While the use of the BSR with an embedded 8-bit address allows users to address the entire range of data memory, it also means that the user must always ensure that the correct bank is selected. Otherwise, data may be read from, or written to, the wrong location. This can be disastrous if a GPR is the intended target of an operation, but an SFR is written to instead. Verifying and/or changing the BSR for each read or write to data memory can become very inefficient.

To streamline access for the most commonly used data memory locations, the data memory is configured with an Access Bank, which allows users to access a mapped block of memory without specifying a BSR. The Access Bank consists of the first 96 bytes of memory (00h-5Fh) in Bank 0 and the last 160 bytes of memory (60h-FFh) in Bank 15. The lower half is known as the "Access RAM" and is composed of GPRs. The upper half is where the device's SFRs are mapped. These two areas are mapped contiguously in the Access Bank and can be addressed in a linear fashion by an 8-bit address (Figure 5-6).

The Access Bank is used by core PIC18 instructions that include the Access RAM bit (the 'a' parameter in the instruction). When 'a' is equal to '1', the instruction uses the BSR and the 8-bit address included in the opcode for the data memory address. When 'a' is '0',

however, the instruction is forced to use the Access Bank address map; the current value of the BSR is ignored entirely.

Using this "forced" addressing allows the instruction to operate on a data address in a single cycle without updating the BSR first. For 8-bit addresses of 60h and above, this means that users can evaluate and operate on SFRs more efficiently. The Access RAM below 60h is a good place for data values that the user might need to access rapidly, such as immediate computational results or common program variables. Access RAM also allows for faster and more code efficient context saving and switching of variables.

The mapping of the Access Bank is slightly different when the extended instruction set is enabled (XINST Configuration bit = 1). This is discussed in more detail in Section 5.6.3 "Mapping the Access Bank in Indexed Literal Offset Mode".

5.3.3 GENERAL PURPOSE REGISTER FILE

PIC18 devices may have banked memory in the GPR area. This is data RAM which is available for use by all instructions. GPRs start at the bottom of Bank 0 (address 000h) and grow upwards towards the bottom of the SFR area. GPRs are not initialized by a Power-on Reset and are unchanged on all other Resets.

5.3.4 SPECIAL FUNCTION REGISTERS

The Special Function Registers (SFRs) are registers used by the CPU and peripheral modules for controlling the desired operation of the device. These registers are implemented as static RAM. SFRs start at the top of data memory (FFFh) and extend downward to occupy more than the top half of Bank 15 (F60h to FFFh). A list of these registers is given in Table 5-2 and Table 5-3. The SFRs can be classified into two sets: those associated with the "core" device functionality (ALU, Resets and interrupts) and those related to the peripheral functions. The Reset and Interrupt registers are described in their respective chapters, while the ALU's STATUS register is described later in this section. Registers related to the operation of the peripheral features are described in the chapter for that peripheral.

The SFRs are typically distributed among the peripherals whose functions they control. Unused SFR locations are unimplemented and read as '0's.

TABLE 5-2: SPECIAL FUNCTION REGISTER MAP FOR PIC18F85J90 FAMILY DEVICES

Address	Name	Address	Name	Address	Name	Address	Name	Address	Name
FFFh	TOSU	FDFh	INDF2 ⁽¹⁾	FBFh	LCDDATA4 ⁽³⁾	F9Fh	IPR1	F7Fh	SPBRGH1
FFEh	TOSH	FDEh	POSTINC2 ⁽¹⁾	FBEh	LCDDATA3	F9Eh	PIR1	F7Eh	BAUDCON1
FFDh	TOSL	FDDh	POSTDEC2 ⁽¹⁾	FBDh	LCDDATA2	F9Dh	PIE1	F7Dh	LCDDATA23 ⁽³⁾
FFCh	STKPTR	FDCh	PREINC2 ⁽¹⁾	FBCh	LCDDATA1	F9Ch	(2)	F7Ch	LCDDATA22 ⁽³⁾
FFBh	PCLATU	FDBh	PLUSW2 ⁽¹⁾	FBBh	LCDDATA0	F9Bh	OSCTUNE	F7Bh	LCDDATA21
FFAh	PCLATH	FDAh	FSR2H	FBAh	LCDSE5 ⁽³⁾	F9Ah	TRISJ ⁽³⁾	F7Ah	LCDDATA20
FF9h	PCL	FD9h	FSR2L	FB9h	LCDSE4 ⁽³⁾	F99h	TRISH ⁽³⁾	F79h	LCDDATA19
FF8h	TBLPTRU	FD8h	STATUS	FB8h	LCDSE3	F98h	TRISG	F78h	LCDDATA18
FF7h	TBLPTRH	FD7h	TMR0H	FB7h	LCDSE2	F97h	TRISF	F77h	LCDDATA17 ⁽³⁾
FF6h	TBLPTRL	FD6h	TMR0L	FB6h	LCDSE1	F96h	TRISE	F76h	LCDDATA16 ⁽³⁾
FF5h	TABLAT	FD5h	TOCON	FB5h	CVRCON	F95h	TRISD	F75h	LCDDATA15
FF4h	PRODH	FD4h	_(2)	FB4h	CMCON	F94h	TRISC	F74h	LCDDATA14
FF3h	PRODL	FD3h	OSCCON	FB3h	TMR3H	F93h	TRISB	F73h	LCDDATA13
FF2h	INTCON	FD2h	LCDREG	FB2h	TMR3L	F92h	TRISA	F72h	LCDDATA12
FF1h	INTCON2	FD1h	WDTCON	FB1h	T3CON	F91h	LATJ ⁽³⁾	F71h	LCDDATA11 ⁽³⁾
FF0h	INTCON3	FD0h	RCON	FB0h	(2)	F90h	LATH ⁽³⁾	F70h	LCDDATA10 ⁽³⁾
FEFh	INDF0 ⁽¹⁾	FCFh	TMR1H	FAFh	SPBRG1	F8Fh	LATG	F6Fh	LCDDATA9
FEEh	POSTINC0 ⁽¹⁾	FCEh	TMR1L	FAEh	RCREG1	F8Eh	LATF	F6Eh	LCDDATA8
FEDh	POSTDEC0 ⁽¹⁾	FCDh	T1CON	FADh	TXREG1	F8Dh	LATE	F6Dh	LCDDATA7
FECh	PREINC0 ⁽¹⁾	FCCh	TMR2	FACh	TXSTA1	F8Ch	LATD	F6Ch	LCDDATA6
FEBh	PLUSW0 ⁽¹⁾	FCBh	PR2	FABh	RCSTA1	F8Bh	LATC	F6Bh	LCDDATA5 ⁽³⁾
FEAh	FSR0H	FCAh	T2CON	FAAh	LCDPS	F8Ah	LATB	F6Ah	CCPR1H
FE9h	FSR0L	FC9h	SSPBUF	FA9h	LCDSE0	F89h	LATA	F69h	CCPR1L
FE8h	WREG	FC8h	SSPADD	FA8h	LCDCON	F88h	PORTJ ⁽³⁾	F68h	CCP1CON
FE7h	INDF1 ⁽¹⁾	FC7h	SSPSTAT	FA7h	EECON2	F87h	PORTH ⁽³⁾	F67h	CCPR2H
FE6h	POSTINC1 ⁽¹⁾	FC6h	SSPCON1	FA6h	EECON1	F86h	PORTG	F66h	CCPR2L
FE5h	POSTDEC1 ⁽¹⁾	FC5h	SSPCON2	FA5h	IPR3	F85h	PORTF	F65h	CCP2CON
FE4h	PREINC1 ⁽¹⁾	FC4h	ADRESH	FA4h	PIR3	F84h	PORTE	F64h	SPBRG2
FE3h	PLUSW1 ⁽¹⁾	FC3h	ADRESL	FA3h	PIE3	F83h	PORTD	F63h	RCREG2
FE2h	FSR1H	FC2h	ADCON0	FA2h	IPR2	F82h	PORTC	F62h	TXREG2
FE1h	FSR1L	FC1h	ADCON1	FA1h	PIR2	F81h	PORTB	F61h	TXSTA2
FE0h	BSR	FC0h	ADCON2	FA0h	PIE2	F80h	PORTA	F60h	RCSTA2

Note 1: This is not a physical register.

2: Unimplemented registers are read as '0'.

3: This register is not available on 64-pin devices.

Filename	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Details on page
TOSU				Top-of-Stack	Upper Byte (1	[[OS<20:16>)			0 0000	51, 59
TOSH	Top-of-Stack High Byte (TOS<15:8>)								0000 0000	51, 59
TOSL	Top-of-Stack Low Byte (TOS<7:0>)								0000 0000	51, 59
STKPTR	STKFUL STKUNF — Return Stack Pointer								uu-0 0000	51,60
PCLATU	_	_	bit 21 ⁽¹⁾		ister for PC<2	0:16>			0 0000	51, 59
PCLATH	Holding Register for PC<15:8>								0000 0000	51, 59
PCL	PC Low Byte (PC<7:0>)								0000 0000	51, 59
TBLPTRU	_	_	bit 21	Program Me	mory Table Po	ointer Upper B	yte (TBLPTR-	<20:16>)	00 0000	51, 84
TBLPTRH	Program Mer	mory Table Pc	ointer High By	te (TBLPTR<1	15:8>)			,	0000 0000	51, 84
TBLPTRL	Program Mer	mory Table Pc	ointer Low Byt	e (TBLPTR<7	:0>)				0000 0000	51, 84
TABLAT		mory Table La		,	,				0000 0000	51, 84
PRODH		ister High Byte							xxxx xxxx	51, 91
PRODL	-	ister Low Byte							xxxx xxxx	51, 91
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INTOIE	RBIE	TMR0IF	INT0IF	RBIF	0000 000x	51,95
INTCON2	RBPU	INTEDG0	INTEDG1	INTEDG2	INTEDG3	TMR0IP	INT3IP	RBIP	1111 1111	51,96
INTCON3	INT2IP	INT1IP	INT3IE	INT2IE	INT1IE	INT3IF	INT2IF	INT1IF	1100 0000	51,97
INDF0	Uses content	ts of FSR0 to	address data	memory – val	ue of FSR0 no	ot changed (no	ot a physical r	egister)	N/A	51, 75
POSTINC0	Uses content	ts of FSR0 to	address data	memory – val	ue of FSR0 po	ost-incremente	ed (not a phys	ical register)	N/A	51, 76
POSTDEC0	Uses contents of FSR0 to address data memory – value of FSR0 post-incremented (not a physical register) Uses contents of FSR0 to address data memory – value of FSR0 post-decremented (not a physical register)							N/A	51, 76	
PREINC0	Uses contents of FSR0 to address data memory – value of FSR0 pre-incremented (not a physical register)							N/A	51, 76	
PLUSW0		ts of FSR0 to a 10 offset by W	address data	memory – valı	ue of FSR0 pro	e-incremented	l (not a physic	al register) –	N/A	51,76
FSR0H	_	_	—	_	Indirect Data	Memory Add	ess Pointer 0	High	xxxx	51, 75
FSR0L	Indirect Data Memory Address Pointer 0 Low Byte								xxxx xxxx	51, 75
WREG	Working Reg	ister							xxxx xxxx	51
INDF1	Uses content	ts of FSR1 to	address data	memory – val	ue of FSR1 no	ot changed (no	ot a physical r	egister)	N/A	51, 75
POSTINC1	Uses content	ts of FSR1 to	address data	memory – val	ue of FSR1 po	ost-incremente	ed (not a phys	ical register)	N/A	51, 76
POSTDEC1	Uses content	ts of FSR1 to	address data	memory – val	ue of FSR1 po	ost-decrement	ed (not a phy	sical register)	N/A	51, 76
PREINC1	Uses content	ts of FSR1 to	address data	memory – val	ue of FSR1 pr	e-incremente	d (not a physi	cal register)	N/A	51, 76
PLUSW1		ts of FSR1 to a 1 offset by W	address data	memory – valı	ue of FSR1 pro	e-incremented	l (not a physic	al register) –	N/A	51,76
FSR1H	_	_	_	_	Indirect Data	Memory Add	ess Pointer 1	High Byte	xxxx	52, 75
FSR1L	Indirect Data	Memory Add	ress Pointer 1	Low Byte					xxxx xxxx	52, 75
BSR	_	_	_	_	Bank Select	Register			0000	52, 64
INDF2	Uses content	ts of FSR2 to	address data	memory – val	ue of FSR2 no	ot changed (no	ot a physical r	egister)	N/A	52, 75
POSTINC2	Uses content	ts of FSR2 to	address data	memory – val	ue of FSR2 po	ost-incremente	ed (not a phys	ical register)	N/A	52, 76
POSTDEC2	Uses content	ts of FSR2 to	address data	memory – val	ue of FSR2 po	ost-decrement	ed (not a phy	sical register)	N/A	52, 76
PREINC2	Uses content	ts of FSR2 to	address data	memory – val	ue of FSR2 pr	e-incremente	d (not a physi	cal register)	N/A	52, 76
PLUSW2		ts of FSR2 to a 2 offset by W	address data	memory – valı	ue of FSR2 pro	e-incremented	l (not a physic	al register) –	N/A	52, 76
FSR2H	—	—	—	_	Indirect Data	Memory Add	ess Pointer 2	High Byte	xxxx	52, 75
FSR2L	Indirect Data	Memory Add	ress Pointer 2	Low Byte					xxxx xxxx	52, 75
STATUS	_	_	_	N	OV	Z	DC	С	x xxxx	52, 73

TABLE 5-3: PIC18F85J90 FAMILY REGISTER FILE SUMMARY

 $\label{eq:logistical_logistical$

Note 1: Bit 21 of the PC is only available in Test mode and Serial Programming modes.

2: These registers and/or bits are available only on 80-pin devices; otherwise they are unimplemented and read as '0'. Reset states shown are for 80-pin devices.

3: Alternate names and definitions for these bits when the MSSP module is operating in I²C™ Slave mode. See Section 16.4.3.2 "Address Masking" for details.

4: The PLLEN bit is only available in specific oscillator configurations; otherwise it is disabled and reads as '0'. See Section 2.4.3 "PLL Frequency Multiplier" for details.

5: RA6/RA7 and their associated latch and direction bits are configured as port pins only when the internal oscillator is selected as the default clock source (FOSC2 Configuration bit = 0); otherwise, they are disabled and these bits read as '0'.

TADLE 3-3	. FICI	01 00030		AWILT REGISTER FILE SUWIWART				NOLD)			
Filename	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Details on page	
TMR0H	Timer0 Register High Byte									52, 133	
TMR0L	Timer0 Regis	ster Low Byte							xxxx xxxx	52, 133	
TOCON	TMR0ON	T08BIT	TOCS	TOSE	PSA	T0PS2	T0PS1	T0PS0	1111 1111	52, 131	
OSCCON	IDLEN	IRCF2	IRCF1	IRCF0	OSTS	IOFS	SCS1	SCS0	0100 q000	30, 52	
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Legend: x = unknown, u = unchanged, - = unimplemented, g = value depends on condition, r = reserved, do not modify

Note 1: Bit 21 of the PC is only available in Test mode and Serial Programming modes.

2: These registers and/or bits are available only on 80-pin devices; otherwise they are unimplemented and read as '0'. Reset states shown are for 80-pin devices.

3: Alternate names and definitions for these bits when the MSSP module is operating in I²C[™] Slave mode. See Section 16.4.3.2 "Address Masking" for details.

4: The PLLEN bit is only available in specific oscillator configurations; otherwise it is disabled and reads as '0'. See Section 2.4.3 "PLL Frequency Multiplier" for details.

5: RA6/RA7 and their associated latch and direction bits are configured as port pins only when the internal oscillator is selected as the default clock source (FOSC2 Configuration bit = 0); otherwise, they are disabled and these bits read as '0'.

RCREG1 EUSART Receive Register 0000 0000 53, 241 TXREG1 EUSART Transmit Register 0000 0001 53, 230 TXSTA1 CSRC TX9 TXEN SPINC SENDB BRGH TRIMT TX9D 0000 0011 53, 230 RCSTA1 SPEN RX9 SSREN CREN ADDEN FERR OERR RX9D 0000 0000 53, 231 LCDFS WFT BIASMD LCDA WA LP3 LP2 LP1 LP0 0000 0000 53, 158 LCDCON LCDEN SLPEN WERR — CS1 CS0 LMUX0 000 000 53, 158 EECON1 — — FREE WRERR WREN WR - 53, 82 EECON1 — LCDIP RC2IP TX2IP — CCP2IP CCP1IP - 11-1-1 54, 100 IPR3 — LCDIP RC2IP TX2IP </th <th>TABLE 5-5.</th> <th></th> <th>01 03030</th> <th></th> <th>LCUDIE</th> <th>n FILE SU</th> <th></th> <th></th> <th></th> <th></th> <th></th>	TABLE 5-5.		01 03030		LCUDIE	n FILE SU					
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PIR1 — ADIF RC1IF TX1IF SSPIF — TMR2IF TMR1IF -000 0-00 54,98 PIE1 — ADIE RC1IE TX1IE SSPIE — TMR2IE TMR1IE -000 0-00 54,98 PIE1 — ADIE RC1IE TX1IE SSPIE — TMR2IE TMR1IE -000 0-00 54,98 OSCTUNE INTSRC PLLEN ⁽⁴⁾ TUN5 TUN4 TUN3 TUN2 TUN1 TUN0 0000 0000 31,54 TRISJ ⁽²⁾ TRISJ7 TRISJ6 TRISJ5 TRISJ4 TRISJ3 TRISJ2 TRISJ0 1111 111 54,120 TRISG SPIOD CCP20D CCP10D TRISG3 TRISG2 TRISG1 TRISG0 0001 111 54,124 TRISF TRISF7 TRISF6 TRISF5 TRISF4 TRISF3 TRISF1 — 1111 111 54,124 TRISF TRISF6	IPR1	_	ADIP	RC1IP	TX1IP	SSPIP	_	TMR2IP	TMR1IP	-111 1-11	
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TRISDTRISD7TRISD6TRISD5TRISD4TRISD3TRISD3TRISD2TRISD1TRISD11111111154, 119TRISCTRISC7TRISC6TRISC5TRISC4TRISC3TRISC2TRISC1TRISC01111111154, 117TRISBTRISB7TRISB6TRISB5TRISB5TRISB4TRISB3TRISB2TRISB1TRISB01111111154, 117TRISATRISA7 ⁽⁵⁾ TRISA6 ⁽⁵⁾ TRISA5TRISA4TRISA3TRISA2TRISA1TRISA01111111154, 111LATJ ⁽²⁾ LATJ7LATJ6LATJ5LATJ4LATJ3LATJ2LATJ1LATJ0xxxx xxxx54, 130LATH ⁽²⁾ LATH7LATH6LATH5LATH4LATH3LATH2LATH1LATG000-x xxxx54, 128LATGU2ODU1ODLATG4LATG3LATE2LATF1xxxx xxxx54, 124LATFLATF7LATF6LATF5LATF4LATF3LATF2LATF1xxxx xxxx54, 124LATELATE7LATE6LATE5LATE4LATE3LATE1LATE0xxxx xxxx54, 119LATDLATD7LAT66LAT5LATE4LATE3LATE1LATG000-x xxxx54, 124LATELATF7LATE6LATE5LATE4LATE3LATE1LATE0xxxx xxxx54, 119LATDLATD7LATC6LATD5LATC4LATD3LA	TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1	_	1111 111-	54, 124
TRISDTRISD7TRISD6TRISD5TRISD4TRISD3TRISD3TRISD2TRISD1TRISD11111111154, 119TRISCTRISC7TRISC6TRISC5TRISC4TRISC3TRISC2TRISC1TRISC01111111154, 117TRISBTRISB7TRISB6TRISB5TRISB5TRISB4TRISB3TRISB2TRISB1TRISB01111111154, 117TRISATRISA7 ⁽⁵⁾ TRISA6 ⁽⁵⁾ TRISA5TRISA4TRISA3TRISA2TRISA1TRISA01111111154, 111LATJ ⁽²⁾ LATJ7LATJ6LATJ5LATJ4LATJ3LATJ2LATJ1LATJ0xxxx xxxx54, 130LATH ⁽²⁾ LATH7LATH6LATH5LATH4LATH3LATH2LATH1LATG000-x xxxx54, 128LATGU2ODU1ODLATG4LATG3LATE2LATF1xxxx xxxx54, 124LATFLATF7LATF6LATF5LATF4LATF3LATF2LATF1xxxx xxxx54, 124LATELATE7LATE6LATE5LATE4LATE3LATE1LATE0xxxx xxxx54, 119LATDLATD7LAT66LAT5LATE4LATE3LATE1LATG000-x xxxx54, 124LATELATF7LATE6LATE5LATE4LATE3LATE1LATE0xxxx xxxx54, 119LATDLATD7LATC6LATD5LATC4LATD3LA	TRISE	TRISE7	TRISE6	TRISE5	TRISE4	TRISE3	_	TRISE1	TRISE0	1111 1-11	54, 121
TRISCTRISC7TRISC6TRISC5TRISC4TRISC3TRISC2TRISC1TRISC1TRISC01111111154, 117TRISBTRISB7TRISB6TRISB5TRISB4TRISB3TRISB3TRISB2TRISB1TRISB01111111154, 114TRISATRISA7 ⁽⁵⁾ TRISA6 ⁽⁵⁾ TRISA5TRISA4TRISA3TRISA2TRISA1TRISA01111111154, 114LATJ ⁽²⁾ LATJ7LATJ6LATJ5LATJ4LATJ3LATJ2LATJ1LATJ0xxxx xxxx54, 130LATH ⁽²⁾ LATH7LATH6LATH5LATH4LATH3LATH2LATH1LATH0xxxx xxxx54, 128LATGU2ODU1OD—LATG4LATG3LATG2LATG1LATG000-x xxxx54, 124LATFLATF7LATF6LATF5LATF4LATF3LATF2LATF1—xxxx xxxx54, 124LATELATD7LATG6LATD5LATE4LATG3LATD2LATD1LATD0xxxx xxxx54, 121LATDLATD7LAT66LATE5LATE4LATE3—LATE1LATE0xxxx xxxx54, 121LATDLATD7LATG6LATC5LATC4LATG3LATD2LATD1LATD0xxxx xxxx54, 121LATDLATD7LATG6LATD5LATC4LATG3LATC2LATD1LATD0xxxx xxxx54, 117LATDLATC7LATC6LATC5LATC4LATC3LATC2<						TRISD3	TRISD2				
TRISBTRISB6TRISB5TRISB4TRISB3TRISB2TRISB1TRISB01111111154, 114TRISATRISA7 ⁽⁵⁾ TRISA6 ⁽⁵⁾ TRISA5TRISA4TRISA3TRISA2TRISA1TRISA01111111154, 114LATJ ⁽²⁾ LATJ7LATJ6LATJ5LATJ4LATJ3LATJ2LATJ1LATJ0xxxx xxxx54, 130LATH ⁽²⁾ LATH7LATH6LATH5LATH4LATH3LATH2LATH1LATH0xxxx xxxx54, 128LATGU2ODU1OD—LATG4LATG3LATG2LATG1LATG000-x xxxx54, 128LATFLATF7LATF6LATF5LATF4LATF3LATF2LATF1—xxxx xxxx54, 124LATELATE7LATE6LATE5LATE4LATE3—LATE1LATE0xxxx xxxx54, 121LATDLATD7LAT66LATE5LATC4LATG3LATD2LATD1LATD0xxxx xxxx54, 121LATCLATC7LATC6LATC5LATC4LATC3LATC2LATC1LATC0xxxx xxxx54, 117LATBLATB7LATB6LATB5LATB4LATB3LATB2LATB1LATB0xxxx xxxx54, 114	TRISC		TRISC6		TRISC4	TRISC3			TRISC0		
LATJ(2)LATJ7LATJ6LATJ5LATJ4LATJ3LATJ2LATJ1LATJ0xxxx xxxx54, 130LATH(2)LATH7LATH6LATH5LATH4LATH3LATH2LATH1LATH0xxxx xxxx54, 130LATGU2ODU1OD—LATG4LATG3LATG2LATG1LATG000-x xxxx54, 128LATFLATF7LATF6LATF5LATF4LATG3LATG2LATG1LATG000-x xxxx54, 126LATFLATF7LATF6LATF5LATF4LATF3LATF2LATF1—xxxx xxxx-54, 124LATELATE7LATE6LATE5LATE4LATE3—LATE1LATE0xxxx xxxx54, 119LATDLATD7LATO6LATD5LATC4LATC3LATC2LATC1LATC0xxxx xxxx54, 117LATBLATB7LATB6LATB5LATB4LATB3LATB2LATB1LATB0xxxx xxxx54, 114	TRISB										
LATH(2)LATH7LATH6LATH5LATH4LATH3LATH2LATH1LATH0xxx xxxx54, 128LATGU2ODU1OD—LATG4LATG3LATG2LATG1LATG000-x xxxx54, 128LATFLATF7LATF6LATF5LATF4LATG3LATG2LATG1LATG000-x xxxx54, 128LATFLATF7LATF6LATF5LATF4LATF3LATF2LATF1—xxxx xxx-54, 124LATELATE7LATE6LATE5LATE4LATE3—LATE1LATE0xxxx xxxx54, 119LATDLATD7LATD6LATD5LATC4LATC3LATC2LATC1LATC0xxxx xxxx54, 117LATBLATB7LATB6LATB5LATB4LATB3LATB2LATB1LATB0xxxx xxxx54, 114	TRISA	TRISA7 ⁽⁵⁾	TRISA6 ⁽⁵⁾	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	1111 1111	54, 111
LATH(2)LATH7LATH6LATH5LATH4LATH3LATH2LATH1LATH0xxx xxxx54, 128LATGU2ODU1OD—LATG4LATG3LATG2LATG1LATG000-x xxxx54, 128LATFLATF7LATF6LATF5LATF4LATG3LATG2LATG1LATG000-x xxxx54, 128LATFLATF7LATF6LATF5LATF4LATF3LATF2LATF1—xxxx xxx-54, 124LATELATE7LATE6LATE5LATE4LATE3—LATE1LATE0xxxx xxxx54, 119LATDLATD7LATD6LATD5LATC4LATC3LATC2LATC1LATC0xxxx xxxx54, 117LATBLATB7LATB6LATB5LATB4LATB3LATB2LATB1LATB0xxxx xxxx54, 114			LATJ6								,
LATGU2ODU1OD—LATG4LATG3LATG2LATG1LATG000-x xxxx54, 126LATFLATF7LATF6LATF5LATF4LATF3LATF2LATF1—xxxx xxxx54, 124LATELATE7LATE6LATE5LATE4LATE3—LATE1LATE0xxxx x-xx54, 121LATDLATD7LATD6LATD5LATD4LATD3LATD2LATD1LATD0xxxx xxxx54, 119LATCLATC7LATC6LATC5LATC4LATC3LATC2LATC1LATC0xxxx xxxx54, 117LATBLATB7LATB6LATB5LATB4LATB3LATB2LATB1LATB0xxxx xxxx54, 114											
LATFLATF6LATF5LATF4LATF3LATF2LATF1—xxx xxx-54, 124LATELATE7LATE6LATE5LATE4LATE3—LATE1LATE0xxx x-xx54, 121LATDLATD7LATD6LATD5LATD4LATD3LATD2LATD1LATD0xxxx x-xx54, 119LATCLATC7LATC6LATC5LATC4LATC3LATC2LATC1LATC0xxxx xxxx54, 117LATBLATB7LATB6LATB5LATB4LATB3LATB2LATB1LATB0xxxx xxxx54, 114	LATG			_							
LATELATE7LATE6LATE5LATE4LATE3—LATE1LATE0xxxx x-xx54, 121LATDLATD7LATD6LATD5LATD4LATD3LATD2LATD1LATD0xxxx xxxx54, 119LATCLATC7LATC6LATC5LATC4LATC3LATC2LATC1LATC0xxxx xxxx54, 117LATBLATB7LATB6LATB5LATB4LATB3LATB2LATB1LATB0xxxx xxxx54, 114				LATF5					_	-	
LATDLATD7LATD6LATD5LATD4LATD3LATD2LATD1LATD0xxxx xxxx54, 119LATCLATC7LATC6LATC5LATC4LATC3LATC2LATC1LATC0xxxx xxxx54, 117LATBLATB7LATB6LATB5LATB4LATB3LATB2LATB1LATB0xxxx xxxx54, 114							_		LATE0		
LATCLATC6LATC5LATC4LATC3LATC2LATC1LATC0xxxx xxxx54, 117LATBLATB7LATB6LATB5LATB4LATB3LATB2LATB1LATB0xxxx xxxx54, 114							LATD2				
LATB LATB7 LATB6 LATB5 LATB4 LATB3 LATB2 LATB1 LATB0 xxxx xxxx 54, 114											-
	LATA	1							LATA0	XXXX XXXX	54, 111

TABLE 5-3:	PIC18F85J90 FAMILY REGISTER FILE SUMMARY ((CONTINUED))

 $\label{eq:logistical_logistical$

Note 1: Bit 21 of the PC is only available in Test mode and Serial Programming modes.

2: These registers and/or bits are available only on 80-pin devices; otherwise they are unimplemented and read as '0'. Reset states shown are for 80-pin devices.

3: Alternate names and definitions for these bits when the MSSP module is operating in I²C[™] Slave mode. See Section 16.4.3.2 "Address Masking" for details.

 The PLLEN bit is only available in specific oscillator configurations; otherwise it is disabled and reads as '0'. See Section 2.4.3 "PLL Frequency Multiplier" for details.

5: RA6/RA7 and their associated latch and direction bits are configured as port pins only when the internal oscillator is selected as the default clock source (FOSC2 Configuration bit = 0); otherwise, they are disabled and these bits read as '0'.

TABLE 5-3:	PIC18F85J90 FAMILY REGISTER FILE SUMMARY	(CONTINUED)
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TADLE J-J.	ABLE 5-3: PICT8F85J90 FAMILY REGISTER FILE SUMMARY (CONTINUED									
Filename	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Details on page
PORTJ ⁽²⁾	RJ7	RJ6	RJ5	RJ4	RJ3	RJ2	RJ1	RJ0	xxxx xxxx	54, 130
PORTH ⁽²⁾	RH7	RH6	RH5	RH4	RH3	RH2	RH1	RH0	xxxx xxxx	54, 128
PORTG	RDPU	REPU	RJPU ⁽²⁾	RG4	RG3	RG2	RG1	RG0	000x xxxx	54, 126
PORTF	RF7	RF6	RF5	RF4	RF3	RF2	RF1	_	xxxx xxx-	54, 124
PORTE	RE7	RE6	RE5	RE4	RE3	_	RE1	RE0	xxxx x-xx	55, 121
PORTD	RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0	xxxx xxxx	55, 119
PORTC	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0	xxxx xxxx	55, 117
PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	xxxx xxxx	55, 114
PORTA	RA7 ⁽⁵⁾	RA6 ⁽⁵⁾	RA5	RA4	RA3	RA2	RA1	RA0	xx0x 0000	55, 111
SPBRGH1	EUSART Bai	ud Rate Gene	rator High By	te					0000 0000	55, 233
BAUDCON1	ABDOVF	RCMT	—	SCKP	BRG16	—	WUE	ABDEN	01-0 0-00	55, 232
LCDDATA23(2)	S47C3	S46C3	S45C3	S44C3	S43C3	S42C3	S41C3	S40C3	XXXX XXXX	55, 161
LCDDATA22	S39C3 ⁽²⁾	S38C3 ⁽²⁾	S37C3 ⁽²⁾	S36C3 ⁽²⁾	S35C3 ⁽²⁾	S34C3 ⁽²⁾	S33C3 ⁽²⁾	S32C3	xxxx xxxx	55, 161
LCDDATA21	S31C3	S30C3	S29C3	S28C3	S27C3	S26C3	S25C3	S24C3	xxxx xxxx	55, 161
LCDDATA20	S23C3	S22C3	S21C3	S20C3	S19C3	S18C3	S17C3	S16C3	xxxx xxxx	55, 161
LCDDATA19	S15C3	S14C3	S13C3	S12C3	S11C3	S10C3	S09C3	S08C3	xxxx xxxx	55, 161
LCDDATA18	S07C3	S06C3	S05C3	S04C3	S03C3	S02C3	S01C3	S00C3	xxxx xxxx	55, 161
LCDDATA17 ⁽²⁾	S47C2	S46C2	S45C2	S44C2	S43C2	S42C2	S41C2	S40C2	xxxx xxxx	55, 161
LCDDATA16	S39C2 ⁽²⁾	S38C2 ⁽²⁾	S37C2 ⁽²⁾	S36C2 ⁽²⁾	S35C2 ⁽²⁾	S34C2 ⁽²⁾	S33C2 ⁽²⁾	S32C2	xxxx xxxx	55, 161
LCDDATA15	S31C2	S30C2	S29C2	S28C2	S27C2	S26C2	S25C2	S24C2	xxxx xxxx	55, 161
LCDDATA14	S23C2	S22C2	S21C2	S20C2	S19C2	S18C2	S17C2	S16C2	xxxx xxxx	55, 161
LCDDATA13	S15C2	S14C2	S13C2	S12C2	S11C2	S10C2	S09C2	S08C2	xxxx xxxx	55, 161
LCDDATA12	S07C2	S06C2	S05C2	S04C2	S03C2	S02C2	S01C2	S00C2	xxxx xxxx	55, 161
LCDDATA11 ⁽²⁾	S47C1	S46C1	S45C1	S44C1	S43C1	S42C1	S41C1	S40C1	xxxx xxxx	55, 161
LCDDATA10	S39C1 ⁽²⁾	S38C1 ⁽²⁾	S37C1 ⁽²⁾	S36C1 ⁽²⁾	S35C1 ⁽²⁾	S34C1 ⁽²⁾	S33C1 ⁽²⁾	S32C1	xxxx xxxx	55, 161
LCDDATA9	S31C1	S30C1	S29C1	S28C1	S27C1	S26C1	S25C1	S24C1	xxxx xxxx	55, 161
LCDDATA8	S23C1	S22C1	S21C1	S20C1	S19C1	S18C1	S17C1	S16C1	xxxx xxxx	55, 161
LCDDATA7	S15C1	S14C1	S13C1	S12C1	S11C1	S10C1	S09C1	S08C1	xxxx xxxx	55, 161
LCDDATA6	S07C1	S06C1	S05C1	S04C1	S03C1	S02C1	S01C1	S00C1	xxxx xxxx	55, 161
LCDDATA5 ⁽²⁾	S47C0	S46C0	S45C0	S44C0	S43C0	S42C0	S41C0	S40C0	xxxx xxxx	55, 161
CCPR1H	Capture/Corr	pare/PWM R	egister 1 High	Byte					xxxx xxxx	55, 148
CCPR1L	Capture/Corr	pare/PWM R	egister 1 Low	Byte					xxxx xxxx	55, 148
CCP1CON	_	_	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	00 0000	55, 147
CCPR2H	Capture/Corr	pare/PWM R	egister 2 High	Byte					xxxx xxxx	55, 148
CCPR2L	Capture/Corr	pare/PWM R	egister 2 Low	Byte					xxxx xxxx	56, 148
CCP2CON	_	—	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	00 0000	56, 147
SPBRG2	AUSART Ba	ud Rate Gene	rator Register						0000 0000	56, 252
RCREG2	AUSART Ree	ceive Register	,						0000 0000	56, 257
TXREG2	AUSART Tra	nsmit Registe	r						0000 0000	56, 255
TXSTA2	CSRC	TX9	TXEN	SYNC	_	BRGH	TRMT	TX9D	0000 -010	56, 250
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000x	56, 251

 $\label{eq:Legend: Legend: Le$

Note 1: Bit 21 of the PC is only available in Test mode and Serial Programming modes.

2: These registers and/or bits are available only on 80-pin devices; otherwise they are unimplemented and read as '0'. Reset states shown are for 80-pin devices.

3: Alternate names and definitions for these bits when the MSSP module is operating in I²C[™] Slave mode. See Section 16.4.3.2 "Address Masking" for details.

4: The PLLEN bit is only available in specific oscillator configurations; otherwise it is disabled and reads as '0'. See Section 2.4.3 "PLL Frequency Multiplier" for details.

5: RA6/RA7 and their associated latch and direction bits are configured as port pins only when the internal oscillator is selected as the default clock source (FOSC2 Configuration bit = 0); otherwise, they are disabled and these bits read as '0'.

5.3.5 STATUS REGISTER

The STATUS register, shown in Register 5-2, contains the arithmetic status of the ALU. The STATUS register can be the operand for any instruction, as with any other register. If the STATUS register is the destination for an instruction that affects the Z, DC, C, OV or N bits, then the write to these five bits is disabled.

These bits are set or cleared according to the device logic. Therefore, the result of an instruction with the STATUS register as destination may be different than intended. For example, CLRF STATUS will set the Z bit but leave the other bits unchanged. The STATUS

register then reads back as '000u u1uu'. It is recommended, therefore, that only BCF, BSF, SWAPF, MOVFF and MOVWF instructions are used to alter the STATUS register because these instructions do not affect the Z, C, DC, OV or N bits in the STATUS register.

For other instructions not affecting any Status bits, see the instruction set summaries in Table 23-2 and Table 23-3.

Note: The C and DC bits operate as a borrow and digit borrow bit respectively, in subtraction.

REGISTER 5-2: STATUS REGISTER

U-0	U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
_	_	_	N	OV	Z	DC ⁽¹⁾	C ⁽²⁾
bit 7							bit (
Legend:							
R = Reada	able bit	W = Writable	bit	U = Unimpler	nented bit, rea	ad as '0'	
-n = Value	at POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown
bit 7-5	Unimpleme	ented: Read as '	0'				
bit 4	N: Negative	bit					
		sed for signed ar LU MSB = 1).	ithmetic (2's co	omplement). It i	ndicates whe	her the result wa	as
	1 = Result v 0 = Result v	vas negative vas positive					
bit 3	OV: Overflo	w bit					
		sed for signed ar ude which cause				verflow of the	
		w occurred for sig	gned arithmetic	c (in this arithme	etic operation)	
bit 2	Z: Zero bit						
		ult of an arithmet ult of an arithmet			o		
bit 1	DC: Digit Ca	arry/Borrow bit ⁽¹⁾					
		ADDLW, SUBL					
		out from the 4th			urred		
bit 0	0 = NO carryC: Carry/Bo	y-out from the 4tl	n low-order bit	of the result			
		ADDLW, SUBL	w and SUBWF i	nstructions:			
		out from the Mos			ccurred		
		y-out from the Me					
Note 1:		polarity is reverse ate (RRF, RLF)					
2:	For borrow, the	, , ,					•

5.4 Data Addressing Modes

Note:	The execution of some instructions in the
	core PIC18 instruction set are changed
	when the PIC18 extended instruction set is
	enabled. See Section 5.6 "Data Memory
	and the Extended Instruction Set" for
	more information.

While the program memory can be addressed in only one way – through the program counter – information in the data memory space can be addressed in several ways. For most instructions, the addressing mode is fixed. Other instructions may use up to three modes, depending on which operands are used and whether or not the extended instruction set is enabled.

The addressing modes are:

- Inherent
- Literal
- Direct
- Indirect

An additional addressing mode, Indexed Literal Offset, is available when the extended instruction set is enabled (XINST Configuration bit = 1). Its operation is discussed in greater detail in **Section 5.6.1 "Indexed Addressing with Literal Offset**".

5.4.1 INHERENT AND LITERAL ADDRESSING

Many PIC18 control instructions do not need any argument at all; they either perform an operation that globally affects the device, or they operate implicitly on one register. This addressing mode is known as Inherent Addressing. Examples include SLEEP, RESET and DAW.

Other instructions work in a similar way, but require an additional explicit argument in the opcode. This is known as Literal Addressing mode, because they require some literal value as an argument. Examples include ADDLW and MOVLW, which respectively, add or move a literal value to the W register. Other examples include CALL and GOTO, which include a 20-bit program memory address.

5.4.2 DIRECT ADDRESSING

Direct Addressing specifies all or part of the source and/or destination address of the operation within the opcode itself. The options are specified by the arguments accompanying the instruction.

In the core PIC18 instruction set, bit-oriented and byte-oriented instructions use some version of Direct Addressing by default. All of these instructions include some 8-bit literal address as their Least Significant Byte. This address specifies either a register address in one of the banks of data RAM (Section 5.3.3 "General Purpose Register File"), or a location in the Access Bank (Section 5.3.2 "Access Bank") as the data source for the instruction. The Access RAM bit 'a' determines how the address is interpreted. When 'a' is '1', the contents of the BSR (Section 5.3.1 "Bank Select Register") are used with the address to determine the complete 12-bit address of the register. When 'a' is '0', the address is interpreted as being a register in the Access Bank. Addressing that uses the Access RAM is sometimes also known as Direct Forced Addressing mode.

A few instructions, such as MOVFF, include the entire 12-bit address (either source or destination) in their opcodes. In these cases, the BSR is ignored entirely.

The destination of the operation's results is determined by the destination bit 'd'. When 'd' is '1', the results are stored back in the source register, overwriting its original contents. When 'd' is '0', the results are stored in the W register. Instructions without the 'd' argument have a destination that is implicit in the instruction; their destination is either the target register being operated on or the W register.

5.4.3 INDIRECT ADDRESSING

Indirect Addressing allows the user to access a location in data memory without giving a fixed address in the instruction. This is done by using File Select Registers (FSRs) as pointers to the locations to be read or written to. Since the FSRs are themselves located in RAM as Special Function Registers, they can also be directly manipulated under program control. This makes FSRs very useful in implementing data structures such as tables and arrays in data memory.

The registers for Indirect Addressing are also implemented with Indirect File Operands (INDFs) that permit automatic manipulation of the pointer value with auto-incrementing, auto-decrementing or offsetting with another value. This allows for efficient code using loops, such as the example of clearing an entire RAM bank in Example 5-5. It also enables users to perform Indexed Addressing and other Stack Pointer operations for program memory in data memory.

EXAMPLE 5-5: HOW TO CLEAR RAM (BANK 1) USING INDIRECT ADDRESSING

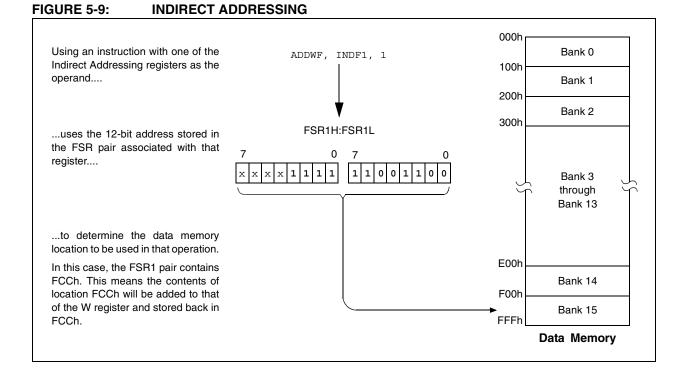
	LFSR	FSR0, 100h	;	
NEXT	CLRF	POSTINC0	;	Clear INDF
			;	register then
			;	inc pointer
	BTFSS	FSROH, 1	;	All done with
			;	Bank1?
	BRA	NEXT	;	NO, clear next
CONTIN	UE		;	YES, continue

5.4.3.1 FSR Registers and the INDF Operand

At the core of Indirect Addressing are three sets of registers: FSR0, FSR1 and FSR2. Each represents a pair of 8-bit registers, FSRnH and FSRnL. The four upper bits of the FSRnH register are not used, so each FSR pair holds a 12-bit value. This represents a value that can address the entire range of the data memory in a linear fashion. The FSR register pairs, then, serve as pointers to data memory locations.

Indirect Addressing is accomplished with a set of Indirect File Operands, INDF0 through INDF2. These can be thought of as "virtual" registers: they are mapped in the SFR space but are not physically implemented. Reading or writing to a particular INDF register actually accesses its corresponding FSR register pair. A read from INDF1, for example, reads the data at the address indicated by FSR1H:FSR1L. Instructions that use the INDF registers as operands actually use the contents of their corresponding FSR as a pointer to the instruction's target. The INDF operand is just a convenient way of using the pointer.

Because Indirect Addressing uses a full 12-bit address, data RAM banking is not necessary. Thus, the current contents of the BSR and the Access RAM bit have no effect on determining the target address.



5.4.3.2 FSR Registers and POSTINC, POSTDEC, PREINC and PLUSW

In addition to the INDF operand, each FSR register pair also has four additional indirect operands. Like INDF, these are "virtual" registers that cannot be indirectly read or written to. Accessing these registers actually accesses the associated FSR register pair, but also performs a specific action on its stored value. They are:

- POSTDEC: accesses the FSR value, then automatically decrements it by '1' afterwards
- POSTINC: accesses the FSR value, then automatically increments it by '1' afterwards
- PREINC: increments the FSR value by '1', then uses it in the operation
- PLUSW: adds the signed value of the W register (range of -127 to 128) to that of the FSR and uses the new value in the operation

In this context, accessing an INDF register uses the value in the FSR registers without changing them. Similarly, accessing a PLUSW register gives the FSR value offset by the value in the W register; neither value is actually changed in the operation. Accessing the other virtual registers changes the value of the FSR registers.

Operations on the FSRs with POSTDEC, POSTINC and PREINC affect the entire register pair; that is, rollovers of the FSRnL register from FFh to 00h carry over to the FSRnH register. On the other hand, results of these operations do not change the value of any flags in the STATUS register (e.g., Z, N, OV, etc.).

The PLUSW register can be used to implement a form of Indexed Addressing in the data memory space. By manipulating the value in the W register, users can reach addresses that are fixed offsets from pointer addresses. In some applications, this can be used to implement some powerful program control structure, such as software stacks, inside of data memory.

5.4.3.3 Operations by FSRs on FSRs

Indirect Addressing operations that target other FSRs or virtual registers represent special cases. For example, using an FSR to point to one of the virtual registers will not result in successful operations. As a specific case, assume that FSR0H:FSR0L contains FE7h, the address of INDF1. Attempts to read the value of the INDF1, using INDF0 as an operand, will return 00h. Attempts to write to INDF1, using INDF0 as the operand, will result in a NOP.

On the other hand, using the virtual registers to write to an FSR pair may not occur as planned. In these cases, the value will be written to the FSR pair but without any incrementing or decrementing. Thus, writing to INDF2 or POSTDEC2 will write the same value to the FSR2H:FSR2L.

Since the FSRs are physical registers mapped in the SFR space, they can be manipulated through all direct operations. Users should proceed cautiously when working on these registers, particularly if their code uses Indirect Addressing.

Similarly, operations by Indirect Addressing are generally permitted on all other SFRs. Users should exercise the appropriate caution that they do not inadvertently change settings that might affect the operation of the device.

5.5 Program Memory and the Extended Instruction Set

The operation of program memory is unaffected by the use of the extended instruction set.

Enabling the extended instruction set adds five additional two-word commands to the existing PIC18 instruction set: ADDFSR, CALLW, MOVSF, MOVSS and SUBFSR. These instructions are executed as described in Section 5.2.4 "Two-Word Instructions".

5.6 Data Memory and the Extended Instruction Set

Enabling the PIC18 extended instruction set (XINST Configuration bit = 1) significantly changes certain aspects of data memory and its addressing. Specifically, the use of the Access Bank for many of the core PIC18 instructions is different; this is due to the introduction of a new addressing mode for the data memory space. This mode also alters the behavior of Indirect Addressing using FSR2 and its associated operands.

What does not change is just as important. The size of the data memory space is unchanged, as well as its linear addressing. The SFR map remains the same. Core PIC18 instructions can still operate in both Direct and Indirect Addressing mode; inherent and literal instructions do not change at all. Indirect Addressing with FSR0 and FSR1 also remains unchanged.

5.6.1 INDEXED ADDRESSING WITH LITERAL OFFSET

Enabling the PIC18 extended instruction set changes the behavior of Indirect Addressing using the FSR2 register pair and its associated file operands. Under the proper conditions, instructions that use the Access Bank – that is, most bit-oriented and byte-oriented instructions – can invoke a form of Indexed Addressing using an offset specified in the instruction. This special addressing mode is known as Indexed Addressing with Literal Offset, or Indexed Literal Offset mode.

When using the extended instruction set, this addressing mode requires the following:

- The use of the Access Bank is forced ('a' = 0); and
- The file address argument is less than or equal to 5Fh.

Under these conditions, the file address of the instruction is not interpreted as the lower byte of an address (used with the BSR in Direct Addressing) or as an 8-bit address in the Access Bank. Instead, the value is interpreted as an offset value to an Address Pointer specified by FSR2. The offset and the contents of FSR2 are added to obtain the target address of the operation.

5.6.2 INSTRUCTIONS AFFECTED BY INDEXED LITERAL OFFSET MODE

Any of the core PIC18 instructions that can use Direct Addressing are potentially affected by the Indexed Literal Offset Addressing mode. This includes all byte-oriented and bit-oriented instructions, or almost one-half of the standard PIC18 instruction set. Instructions that only use Inherent or Literal Addressing modes are unaffected.

Additionally, byte-oriented and bit-oriented instructions are not affected if they use the Access Bank (Access RAM bit is '1') or include a file address of 60h or above. Instructions meeting these criteria will continue to execute as before. A comparison of the different possible addressing modes when the extended instruction set is enabled is shown in Figure 5-10.

Those who desire to use byte-oriented or bit-oriented instructions in the Indexed Literal Offset mode should note the changes to assembler syntax for this mode. This is described in more detail in **Section 23.2.1** "Extended Instruction Syntax".

FIGURE 5-10: COMPARING ADDRESSING OPTIONS FOR BIT-ORIENTED AND BYTE-ORIENTED INSTRUCTIONS (EXTENDED INSTRUCTION SET ENABLED)

EXAMPLE INSTRUCTION: ADDWF, f, d, a (Opcode: 0010 01da ffff ffff)

When a = 0 and $f \ge 60h$:

The instruction executes in Direct Forced mode. 'f' is interpreted as a location in the Access RAM between 060h and FFFh. This is the same as locations F60h to FFFh (Bank 15) of data memory.

Locations below 060h are not available in this addressing mode.

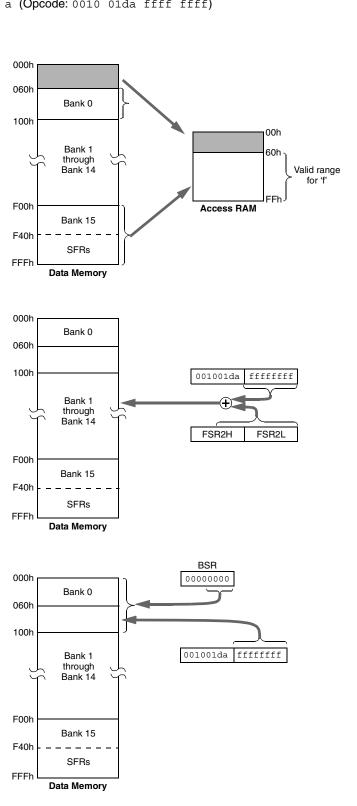
When a = 0 and $f \le 5Fh$:

The instruction executes in Indexed Literal Offset mode. 'f' is interpreted as an offset to the address value in FSR2. The two are added together to obtain the address of the target register for the instruction. The address can be anywhere in the data memory space.

Note that in this mode, the correct syntax is now: ADDWF [k], d where 'k' is the same as 'f'.

When a = 1 (all values of f):

The instruction executes in Direct mode (also known as Direct Long mode). 'f' is interpreted as a location in one of the 16 banks of the data memory space. The bank is designated by the Bank Select Register (BSR). The address can be in any implemented bank in the data memory space.



5.6.3 MAPPING THE ACCESS BANK IN INDEXED LITERAL OFFSET MODE

The use of Indexed Literal Offset Addressing mode effectively changes how the lower part of Access RAM (00h to 5Fh) is mapped. Rather than containing just the contents of the bottom part of Bank 0, this mode maps the contents from Bank 0 and a user-defined "window" that can be located anywhere in the data memory space. The value of FSR2 establishes the lower boundary of the addresses mapped into the window, while the upper boundary is defined by FSR2 plus 95 (5Fh). Addresses in the Access RAM above 5Fh are mapped as previously described (see Section 5.3.2 "Access Bank"). An example of Access Bank remapping in this addressing mode is shown in Figure 5-11.

Remapping of the Access Bank applies *only* to operations using the Indexed Literal Offset mode. Operations that use the BSR (Access RAM bit is '1') will continue to use Direct Addressing as before. Any Indirect or Indexed Addressing operation that explicitly uses any of the indirect file operands (including FSR2) will continue to operate as standard Indirect Addressing. Any instruction that uses the Access Bank, but includes a register address of greater than 05Fh, will use Direct Addressing and the normal Access Bank map.

5.6.4 BSR IN INDEXED LITERAL OFFSET MODE

Although the Access Bank is remapped when the extended instruction set is enabled, the operation of the BSR remains unchanged. Direct Addressing, using the BSR to select the data memory bank, operates in the same manner as previously described.

FIGURE 5-11: REMAPPING THE ACCESS BANK WITH INDEXED LITERAL OFFSET ADDRESSING

Example Situation: 000h ADDWF f, d, a Not Accessible 05Fh FSR2H:FSR2L = 120hBank 0 Locations in the region from the FSR2 Pointer 100h (120h) to the pointer plus 120h Window 05Fh (17Fh) are mapped 17Fh 00h to the bottom of the Bank 1 Bank 1 "Window" Access RAM (000h-05Fh). 200h 5Fh Special Function Regis-60h ters at F60h through FFFh are mapped to 60h Bank 2 through FFh, as usual. SFRs through Bank 14 Bank 0 addresses below 5Fh are not available in FFh this mode. They can still Access Bank be addressed by using the F00h BSR. Bank 15 F60h SFRs FFFh **Data Memory**

NOTES:

6.0 FLASH PROGRAM MEMORY

The Flash program memory is readable, writable and erasable during normal operation over the entire VDD range.

A read from program memory is executed on one byte at a time. A write to program memory is executed on blocks of 64 bytes at a time. Program memory is erased in blocks of 1024 bytes at a time. A Bulk Erase operation may not be issued from user code.

Writing or erasing program memory will cease instruction fetches until the operation is complete. The program memory cannot be accessed during the write or erase, therefore, code cannot execute. An internal programming timer terminates program memory writes and erases.

A value written to program memory does not need to be a valid instruction. Executing a program memory location that forms an invalid instruction results in a NOP.

6.1 Table Reads and Table Writes

In order to read and write program memory, there are two operations that allow the processor to move bytes between the program memory space and the data RAM:

- Table Read (TBLRD)
- Table Write (TBLWT)

The program memory space is 16 bits wide, while the data RAM space is 8 bits wide. Table reads and table writes move data between these two memory spaces through an 8-bit register (TABLAT).

Table read operations retrieve data from program memory and place it into the data RAM space. Figure 6-1 shows the operation of a table read with program memory and data RAM.

Table write operations store data from the data memory space into holding registers in program memory. The procedure to write the contents of the holding registers into program memory is detailed in **Section 6.5** "**Writing to Flash Program Memory**". Figure 6-2 shows the operation of a table write with program memory and data RAM.

Table operations work with byte entities. A table block containing data, rather than program instructions, is not required to be word-aligned. Therefore, a table block can start and end at any byte address. If a table write is being used to write executable code into program memory, program instructions will need to be word-aligned.

FIGURE 6-1: TABLE READ OPERATION

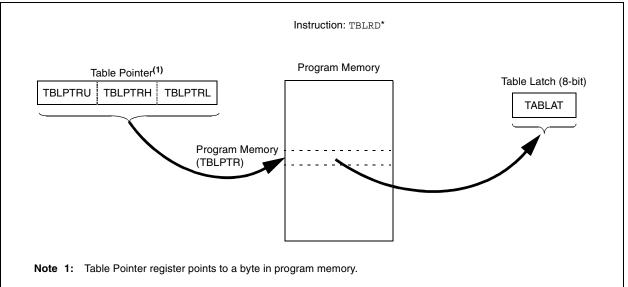
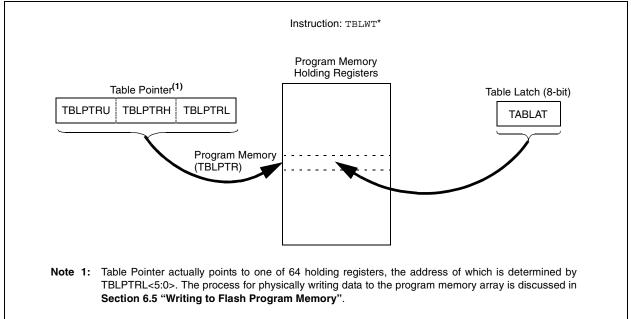


FIGURE 6-2: TABLE WRITE OPERATION



6.2 Control Registers

Several control registers are used in conjunction with the ${\tt TBLRD}$ and ${\tt TBLWT}$ instructions. These include the:

- EECON1 register
- EECON2 register
- TABLAT register
- TBLPTR registers

6.2.1 EECON1 AND EECON2 REGISTERS

The EECON1 register (Register 6-1) is the control register for memory accesses. The EECON2 register is not a physical register; it is used exclusively in the memory write and erase sequences. Reading EECON2 will read all '0's.

The FREE bit, when set, will allow a program memory erase operation. When FREE is set, the erase operation is initiated on the next WR command. When FREE is clear, only writes are enabled. The WREN bit, when set, will allow a write operation. On power-up, the WREN bit is clear. The WRERR bit is set in hardware when the WR bit is set and cleared when the internal programming timer expires and the write operation is complete.

Note:	During normal operation, the WRERR is
	read as '1'. This can indicate that a write
	operation was prematurely terminated by
	a Reset, or a write operation was
	attempted improperly.

The WR control bit initiates write operations. The bit cannot be cleared, only set, in software. It is cleared in hardware at the completion of the write operation.

REGISTER 6-1: EECON1: EEPROM CONTROL REGISTER 1

U-0	U-0	U-0	R/W-0	R/W-x	R/W-0	R/S-0	U-0
—	—	—	FREE	WRERR	WREN	WR	—
bit 7							bit 0

Legend:	U = Unimplemented bi	U = Unimplemented bit, read as '0'			
R = Readable bit	W = Writable bit	S = Set only bit (cannot	be cleared in software)		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown		

bit 7-5	Unimplemented: Read as '0'
bit 4	FREE: Flash Row Erase Enable bit
	1 = Erase the program memory row addressed by TBLPTR on the next WR command (cleared by completion of erase operation)
	0 = Perform write only
bit 3	WRERR: Flash Program Error Flag bit
	 1 = A write operation is prematurely terminated (any Reset during self-timed programming in normal operation, or an improper write attempt) 0 = The write operation completed
bit 2	WREN: Flash Program Write Enable bit
	 1 = Allows write cycles to Flash program memory 0 = Inhibits write cycles to Flash program memory
bit 1	WR: Write Control bit
	 1 = Initiates a program memory erase cycle or write cycle (The operation is self-timed and the bit is cleared by hardware once write is complete. The WR bit can only be set (not cleared) in software.) 0 = Write cycle is complete
bit 0	Unimplemented: Read as '0'

6.2.2 TABLE LATCH REGISTER (TABLAT)

The Table Latch (TABLAT) is an 8-bit register mapped into the SFR space. The Table Latch register is used to hold 8-bit data during data transfers between program memory and data RAM.

6.2.3 TABLE POINTER REGISTER (TBLPTR)

The Table Pointer (TBLPTR) register addresses a byte within the program memory. The TBLPTR is comprised of three SFR registers: Table Pointer Upper Byte, Table Pointer High Byte and Table Pointer Low Byte (TBLPTRU:TBLPTRH:TBLPTRL). These three registers join to form a 22-bit wide pointer. The low-order 21 bits allow the device to address up to 2 Mbytes of program memory space. The 22nd bit allows access to the device ID, the user ID and the Configuration bits.

The Table Pointer register, TBLPTR, is used by the TBLRD and TBLWT instructions. These instructions can update the TBLPTR in one of four ways based on the table operation. These operations are shown in Table 6-1. These operations on the TBLPTR only affect the low-order 21 bits.

6.2.4 TABLE POINTER BOUNDARIES

TBLPTR is used in reads, writes and erases of the Flash program memory.

When a TBLRD is executed, all 22 bits of the TBLPTR determine which byte is read from program memory into TABLAT.

When a TBLWT is executed, the seven LSbs of the Table Pointer register (TBLPTR<6:0>) determine which of the 64 program memory holding registers is written to. When the timed write to program memory begins (via the WR bit), the 12 MSbs of the TBLPTR (TBLPTR<21:10>) determine which program memory block of 1024 bytes is written to. For more detail, see **Section 6.5 "Writing to Flash Program Memory"**.

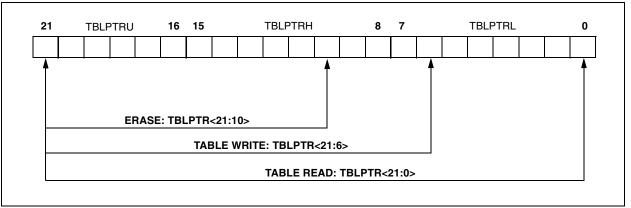
When an erase of program memory is executed, the 12 MSbs of the Table Pointer register point to the 1024-byte block that will be erased. The Least Significant bits are ignored.

Figure 6-3 describes the relevant boundaries of the TBLPTR based on Flash program memory operations.

TABLE 6-1: TABLE POINTER OPERATIONS WITH TBLRD AND TBLWT INSTRUCTIONS

Example	Operation on Table Pointer
TBLRD* TBLWT*	TBLPTR is not modified
TBLRD*+ TBLWT*+	TBLPTR is incremented after the read/write
TBLRD*- TBLWT*-	TBLPTR is decremented after the read/write
TBLRD+* TBLWT+*	TBLPTR is incremented before the read/write

FIGURE 6-3: TABLE POINTER BOUNDARIES BASED ON OPERATION

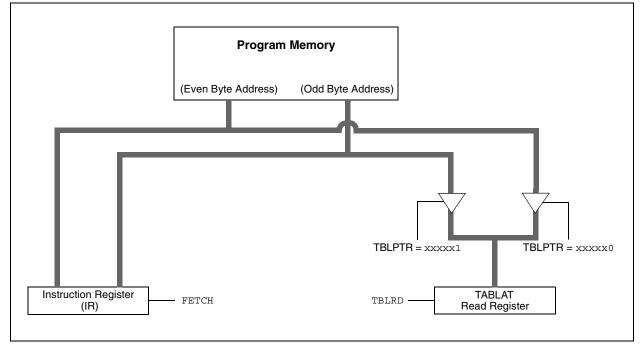


6.3 Reading the Flash Program Memory

The TBLRD instruction is used to retrieve data from program memory and places it into data RAM. Table reads from program memory are performed one byte at a time.

TBLPTR points to a byte address in program space. Executing TBLRD places the byte pointed to into TABLAT. In addition, TBLPTR can be modified automatically for the next table read operation.

FIGURE 6-4: READS FROM FLASH PROGRAM MEMORY



EXAMPLE 6-1: READING A FLASH PROGRAM MEMORY WORD

	MOVLW MOVWF MOVLW MOVWF MOVLW MOVWF	CODE_ADDR_UPPER TBLPTRU CODE_ADDR_HIGH TBLPTRH CODE_ADDR_LOW TBLPTRL	; Load TBLPTR with the base ; address of the word
READ WORD	HOVWI		
	TBLRD*+ MOVF MOVWF TBLRD*+ MOVF MOVWF	TABLAT, W WORD_EVEN TABLAT, W WORD_ODD	; read into TABLAT and increment ; get data ; read into TABLAT and increment ; get data

6.4 Erasing Flash Program Memory

The minimum erase block is 512 words or 1024 bytes. Only through the use of an external programmer, or through ICSP control, can larger blocks of program memory be Bulk Erased. Word erase in the Flash array is not supported.

When initiating an erase sequence from the microcontroller itself, a block of 1024 bytes of program memory is erased. The Most Significant 12 bits of the TBLPTR<21:10> point to the block being erased; TBLPTR<9:0> are ignored.

The EECON1 register commands the erase operation. The WREN bit must be set to enable write operations. The FREE bit is set to select an erase operation. For protection, the write initiate sequence for EECON2 must be used.

A long write is necessary for erasing the internal Flash. Instruction execution is halted while in a long write cycle. The long write will be terminated by the internal programming timer.

6.4.1 FLASH PROGRAM MEMORY ERASE SEQUENCE

The sequence of events for erasing a block of internal program memory location is:

- 1. Load Table Pointer register with address of row being erased.
- 2. Set the WREN and FREE bits (EECON1<2,4>) to enable the erase operation.
- 3. Disable interrupts.
- 4. Write 55h to EECON2.
- 5. Write 0AAh to EECON2.
- 6. Set the WR bit. This will begin the Row Erase cycle.
- The CPU will stall for duration of the erase for Tiw (see parameter D133A).
- 8. Re-enable interrupts.

EXAMPLE 6-2: ERASING A FLASH PROGRAM MEMORY ROW

	MOVLW MOVWF MOVLW MOVWF	CODE_ADDR_UPPER TBLPTRU CODE_ADDR_HIGH TBLPTRH	; load TBLPTR with the base ; address of the memory block
EDACE DOM	MOVLW MOVWF	CODE_ADDR_LOW TBLPTRL	
ERASE_ROW	BSF BSF BCF	EECON1, WREN EECON1, FREE INTCON, GIE	; enable write to memory ; enable Row Erase operation ; disable interrupts
Required Sequence	MOVLW MOVWF MOVLW MOVWF BSF	55h EECON2 0AAh EECON2 EECON1, WR	; write 55h ; write 0AAh ; start erase (CPU stall)
	BSF	INTCON, GIE	; re-enable interrupts

6.5 Writing to Flash Program Memory

The minimum programming block is 32 words or 64 bytes. Word or byte programming is not supported.

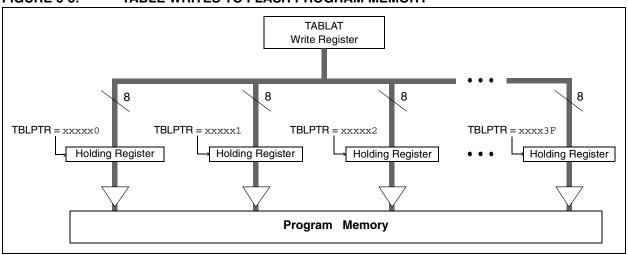
Table writes are used internally to load the holding registers needed to program the Flash memory. There are 64 holding registers used by the table writes for programming.

Since the Table Latch (TABLAT) is only a single byte, the TBLWT instruction may need to be executed 64 times for each programming operation. All of the table write operations will essentially be short writes because only the holding registers are written. At the end of updating the 64 holding registers, the EECON1 register must be written to in order to start the programming operation with a long write.

The long write is necessary for programming the internal Flash. Instruction execution is halted while in a long write cycle. The long write will be terminated by the internal programming timer. The on-chip timer controls the write time. The write/erase voltages are generated by an on-chip charge pump, rated to operate over the voltage range of the device.

- Note 1: Unlike previous PIC[®] devices, members of the PIC18F85J90 family do not reset the holding registers after a write occurs. The holding registers must be cleared or overwritten before a programming sequence.
 - 2: To maintain the endurance of the program memory cells, each Flash byte should not be programmed more than one time between erase operations. Before attempting to modify the contents of the target cell a second time, a Row Erase of the target row, or a Bulk Erase of the entire memory, must be performed.





6.5.1 FLASH PROGRAM MEMORY WRITE SEQUENCE

The sequence of events for programming an internal program memory location should be:

- 1. Read 1024 bytes into RAM.
- 2. Update data values in RAM as necessary.
- 3. Load Table Pointer register with address being erased.
- 4. Execute the Row Erase procedure.
- 5. Load Table Pointer register with address of first byte being written, minus 1.
- 6. Write the 64 bytes into the holding registers with auto-increment.
- 7. Set the WREN bit (EECON1<2>) to enable byte writes.

- 8. Disable interrupts.
- 9. Write 55h to EECON2.
- 10. Write 0AAh to EECON2.
- 11. Set the WR bit. This will begin the write cycle.
- 12. The CPU will stall for duration of the write for Tiw (see parameter D133A).
- 13. Re-enable interrupts.
- 14. Repeat steps 6 through 13 until all 1024 bytes are written to program memory.
- 15. Verify the memory (table read).

An example of the required code is shown in Example 6-3 on the following page.

Note: Before setting the WR bit, the Table Pointer address needs to be within the intended address range of the 64 bytes in the holding register.

	MOVLW MOVWF	CODE_ADDR_UPPER	; Load TBLPTR with the base address
	MOVWF		
		TBLPTRU	; of the memory block, minus 1
	MOVLW	CODE ADDR HIGH	, , ,
	MOVWF	TBLPTRH	
	MOVLW	CODE ADDR LOW	
	MOVWF	TBLPTRL	
ERASE BLOCK	110 0 011		
DIGOD_DIGON	BSF	EECON1, WREN	; enable write to memory
	BSF	EECON1, FREE	; enable Row Erase operation
	BCF	INTCON, GIE	; disable interrupts
	MOVLW	55h	; disable incellupts
	MOVUW	EECON2	; write 55h
			; write 55H
	MOVLW	0AAh EECOND	write 077h
	MOVWF	EECON2	; write 0AAh
	BSF	EECON1, WR	; start erase (CPU stall)
	BSF	INTCON, GIE	; re-enable interrupts
	MOVLW	D'16'	
	MOVWF	WRITE_COUNTER	; Need to write 16 blocks of 64 to write
			; one erase block of 1024
RESTART_BUFFER			
	MOVLW	D'64'	
	MOVWF	COUNTER	
	MOVLW	BUFFER_ADDR_HIGH	; point to buffer
	MOVWF	FSROH	
	MOVLW	BUFFER_ADDR_LOW	
	MOVWF	FSROL	
FILL_BUFFER			
			; read the new data from I2C, SPI,
			; PSP, USART, etc.
WRITE_BUFFER			
	MOVLW	D'64	; number of bytes in holding register
	MOVWF	COUNTER	
WRITE BYTE TO HRE	GS		
	MOVFF	POSTINCO, WREG	; get low byte of buffer data
	MOVWF	TABLAT	; present data to table latch
	TBLWT+*		; write data, perform a short write
			; to internal TBLWT holding register.
	DECFSZ	COUNTER	; loop until buffers are full
	BRA	WRITE BYTE TO HREGS	· •
PROGRAM MEMORY	*		
	BSF	EECON1, WREN	; enable write to memory
	BCF	INTCON, GIE	; disable interrupts
	MOVLW	55h	, albabic incertapes
Required	MOVLW MOVWF	EECON2	; write 55h
-		0AAh	/ WIICE JJH
Sequence	MOVLW		write 077h
	MOVWF	EECON2	; write OAAh
	BSF	EECON1, WR	; start program (CPU stall)
	BSF	INTCON, GIE	; re-enable interrupts
	BCF	EECON1, WREN	; disable write to memory
	DECFSZ	WRITE_COUNTER	; done with one write cycle
	BRA	RESTART_BUFFER	; if not done replacing the erase block

EXAMPLE 6-3: WRITING TO FLASH PROGRAM MEMORY

6.5.2 WRITE VERIFY

Depending on the application, good programming practice may dictate that the value written to the memory should be verified against the original value. This should be used in applications where excessive writes can stress bits near the specification limit.

6.5.3 UNEXPECTED TERMINATION OF WRITE OPERATION

If a write is terminated by an unplanned event, such as loss of power or an unexpected Reset, the memory location just programmed should be verified and reprogrammed if needed. If the write operation is interrupted by a MCLR Reset or a WDT Time-out Reset during normal operation, the user can check the WRERR bit and rewrite the location(s) as needed.

6.6 Flash Program Operation During Code Protection

See Section 22.6 "Program Verification and Code Protection" for details on code protection of Flash program memory.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
TBLPTRU	—	—	bit 21	Program Me	mory Table F	Pointer Upper	Byte (TBLP	TR<20:16>)	51
TBPLTRH	Program M	emory Table	Pointer H	ligh Byte (TE	BLPTR<15:8	>)			51
TBLPTRL	Program M	emory Table	Pointer L	ow Byte (TB	LPTR<7:0>)				51
TABLAT	Program M	Program Memory Table Latch					51		
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INTOIE	RBIE	TMR0IF	INT0IF	RBIF	51
EECON2	EEPROM (Control Regi	ster 2 (not	a physical r	egister)				53
EECON1	_	_	_	FREE	WRERR	WREN	WR	_	53

TABLE 6-2: REGISTERS ASSOCIATED WITH PROGRAM FLASH MEMORY

Legend: — = unimplemented, read as '0'. Shaded cells are not used during program memory access.

NOTES:

7.0 8 x 8 HARDWARE MULTIPLIER

7.1 Introduction

All PIC18 devices include an 8 x 8 hardware multiplier as part of the ALU. The multiplier performs an unsigned operation and yields a 16-bit result that is stored in the Product register pair, PRODH:PRODL. The multiplier's operation does not affect any flags in the STATUS register.

Making multiplication a hardware operation allows it to be completed in a single instruction cycle. This has the advantages of higher computational throughput and reduced code size for multiplication algorithms and allows the PIC18 devices to be used in many applications previously reserved for digital signal processors. A comparison of various hardware and software multiply operations, along with the savings in memory and execution time, is shown in Table 7-1.

7.2 Operation

Example 7-1 shows the instruction sequence for an 8×8 unsigned multiplication. Only one instruction is required when one of the arguments is already loaded in the WREG register.

Example 7-2 shows the sequence to do an 8 x 8 signed multiplication. To account for the sign bits of the arguments, each argument's Most Significant bit (MSb) is tested and the appropriate subtractions are done.

EXAMPLE 7-1: 8 x 8 UNSIGNED MULTIPLY ROUTINE

MOVF	ARG1,	W	;	
MULWF	ARG2		;	ARG1 * ARG2 ->
			;	PRODH: PRODL

EXAMPLE 7-2: 8 x 8 SIGNED MULTIPLY ROUTINE

		HOUTHILE	
MOVF	ARG1, W		
MULWF	ARG2	; ARG1 * ARG2 ->	
		; PRODH:PRODL	
BTFSC	ARG2, SB	; Test Sign Bit	
SUBWF	PRODH, F	; PRODH = PRODH	
		; - ARG1	
MOVF	ARG2, W		
BTFSC	ARG1, SB	; Test Sign Bit	
SUBWF	PRODH, F	; PRODH = PRODH	
		; - ARG2	

		Program	Cycles (Max)	Time			
Routine	Multiply Method	Memory (Words)		@ 40 MHz	@ 10 MHz	@ 4 MHz	
	Without hardware multiply	13	69	6.9 μs	27.6 μs	69 µs	
8 x 8 unsigned	Hardware multiply	1	1	100 ns	400 ns	1 μs	
	Without hardware multiply	33	91	9.1 μs	36.4 μs	91 μs	
8 x 8 signed	Hardware multiply	6	6	600 ns	2.4 μs	6 μs	
16 v 16 uppigned	Without hardware multiply	21	242	24.2 μs	96.8 µs	242 μs	
16 x 16 unsigned	Hardware multiply	28	28	2.8 μs	11.2 μs	28 µs	
16 v 16 signad	Without hardware multiply	52	254	25.4 μs	102.6 μs	254 μs	
16 x 16 signed	Hardware multiply	35	40	4.0 μs	16.0 μs	40 µs	

TABLE 7-1: PERFORMANCE COMPARISON FOR VARIOUS MULTIPLY OPERATIONS

Example 7-3 shows the sequence to do a 16 x 16 unsigned multiplication. Equation 7-1 shows the algorithm that is used. The 32-bit result is stored in four registers (RES3:RES0).

EQUATION 7-1: 16 x 16 UNSIGNED MULTIPLICATION ALGORITHM

RES3:RES0	=	ARG1H:ARG1L • ARG2H:ARG2L
	=	$(ARG1H \bullet ARG2H \bullet 2^{16}) +$
		$(ARG1H \bullet ARG2L \bullet 2^8) +$
		$(ARG1L \bullet ARG2H \bullet 2^8) +$
		$(ARG1L \bullet ARG2L)$

EXAMPLE 7-3:

16 x 16 UNSIGNED MULTIPLY ROUTINE

	MOVF	ARG1L, W	
	MULWF	ARG2L	; ARG1L * ARG2L->
			; PRODH:PRODL
	MOVFF	PRODH, RES1	;
	MOVFF	PRODL, RESO	;
;			
	MOVF	ARG1H, W	
	MULWF	ARG2H	; ARG1H * ARG2H->
			; PRODH:PRODL
	MOVFF	PRODH, RES3	;
	MOVFF	PRODL, RES2	;
;			
	MOVF	ARG1L, W	
	MULWF	ARG2H	; ARG1L * ARG2H->
			; PRODH:PRODL
	MOVF		;
	ADDWF	RES1, F	; Add cross
		PRODH, W	; products
	ADDWFC	RES2, F	;
	CLRF	WREG	;
	ADDWFC	RES3, F	;
;			
	MOVF	ARG1H, W	;
	MULWF	ARG2L	; ARG1H * ARG2L->
			; PRODH:PRODL
	MOVF		;
	ADDWF	RES1, F	; Add cross
	MOVF		; products
		RES2, F	;
		WREG	;
	ADDWFC	RES3, F	;

Example 7-4 shows the sequence to do a 16 x 16 signed multiply. Equation 7-2 shows the algorithm used. The 32-bit result is stored in four registers (RES3:RES0). To account for the sign bits of the arguments, the MSb for each argument pair is tested and the appropriate subtractions are done.

EQUATION 7-2: 16 x 16 SIGNED MULTIPLICATION ALGORITHM

RES3:RES0=ARG1H:ARG1L • ARG2H:ARG2L
$= (ARG1H \bullet ARG2H \bullet 2^{16}) +$
$(ARG1H \bullet ARG2L \bullet 2^8) +$
$(ARG1L \bullet ARG2H \bullet 2^8) +$
$(ARG1L \bullet ARG2L) +$
$(-1 \bullet ARG2H < 7 > \bullet ARG1H: ARG1L \bullet 2^{16}) +$
$(-1 \bullet ARG1H < 7 > \bullet ARG2H: ARG2L \bullet 2^{16})$

EXAMPLE 7-4: 16 x 16 SIGNED MULTIPLY ROUTINE

	WOLI	
MOVF	ARG1L, W	
MULWF	ARG2L	; ARG1L * ARG2L ->
		; PRODH:PRODL
MOVFF	PRODH, RES1	;
MOVFF	PRODL, RESO	
;		,
MOVF	ARG1H, W	
	ARG2H	; ARG1H * ARG2H ->
HOLWI	AIGZII	; PRODH:PRODL
MOVEE	PRODH, RES3	
MOVEF		
	PRODL, RES2	;
; MOME	ADCII W	
MOVF	ARG1L, W	
MOTME	ARG2H	; ARG1L * ARG2H ->
		; PRODH:PRODL
	PRODL, W	;
ADDWF	RES1, F	; Add cross
	PRODH, W	; products
	RES2, F	;
CLRF		;
ADDWFC	RES3, F	;
;		
	ARG1H, W	;
MULWF	ARG2L	; ARG1H * ARG2L ->
		; PRODH:PRODL
MOVF	PRODL, W	;
ADDWF	RES1, F	; Add cross
MOVF	PRODH, W	; products
ADDWFC	RES2, F	;
CLRF	WREG	;
ADDWFC	RES3, F	;
;		
BTFSS	ARG2H, 7	; ARG2H:ARG2L neg?
BRA	SIGN ARG1	; no, check ARG1
MOVF	ARG1L, W	i
SUBWF	RES2	;
	ARG1H, W	;
SUBWFB		
;		
, SIGN ARG1		
	ARG1H, 7	; ARG1H:ARG1L neg?
BRA	CONT CODE	; no, done
	ARG2L, W	;
SUBWF	RES2	
	ARG2H, W	;
SUBWFB		,
	1000	
; CONT CODE		
CONT_CODE		
:		

8.0 INTERRUPTS

Members of the PIC18F85J90 family of devices have multiple interrupt sources and an interrupt priority feature that allows most interrupt sources to be assigned a high priority level or a low priority level. The high priority interrupt vector is at 0008h and the low priority interrupt vector is at 0018h. High priority interrupt events will interrupt any low priority interrupts that may be in progress.

There are thirteen registers which are used to control interrupt operation. These registers are:

- RCON
- INTCON
- INTCON2
- INTCON3
- PIR1, PIR2, PIR3
- PIE1, PIE2, PIE3
- IPR1, IPR2, IPR3

It is recommended that the Microchip header files supplied with MPLAB[®] IDE be used for the symbolic bit names in these registers. This allows the assembler/compiler to automatically take care of the placement of these bits within the specified register.

In general, interrupt sources have three bits to control their operation. They are:

- Flag bit to indicate that an interrupt event occurred
- Enable bit that allows program execution to branch to the interrupt vector address when the flag bit is set
- **Priority bit** to select high priority or low priority

The interrupt priority feature is enabled by setting the IPEN bit (RCON<7>). When interrupt priority is enabled, there are two bits which enable interrupts globally. Setting the GIEH bit (INTCON<7>) enables all interrupts that have the priority bit set (high priority). Setting the GIEL bit (INTCON<6>) enables all interrupts that have the priority bit cleared (low priority). When the interrupt flag, enable bit and appropriate global interrupt enable bit are set, the interrupt will vector immediately to address 0008h or 0018h, depending on the priority bit setting. Individual interrupts can be disabled through their corresponding enable bits.

When the IPEN bit is cleared (default state), the interrupt priority feature is disabled and interrupts are compatible with PIC[®] mid-range devices. In Compatibility mode, the interrupt priority bits for each source have no effect. INTCON<6> is the PEIE bit which enables/disables all peripheral interrupt sources. INTCON<7> is the GIE bit which enables/disables all interrupt sources. All interrupts branch to address 0008h in Compatibility mode.

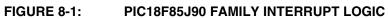
When an interrupt is responded to, the global interrupt enable bit is cleared to disable further interrupts. If the IPEN bit is cleared, this is the GIE bit. If interrupt priority levels are used, this will be either the GIEH or GIEL bit. High priority interrupt sources can interrupt a low priority interrupt. Low priority interrupts are not processed while high priority interrupts are in progress.

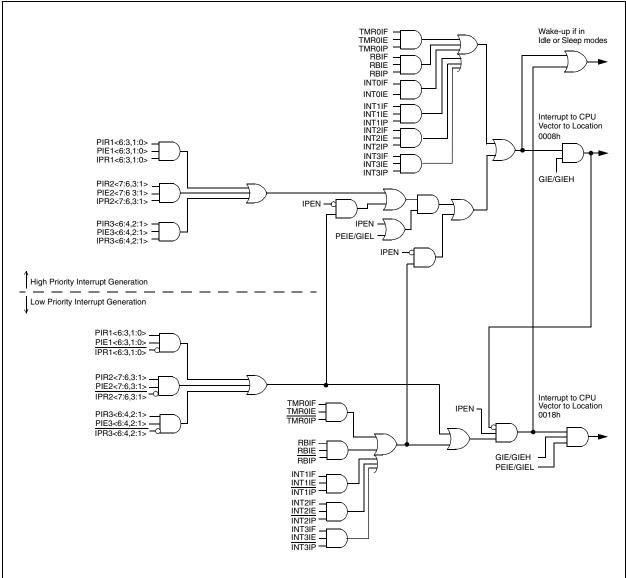
The return address is pushed onto the stack and the PC is loaded with the interrupt vector address (0008h or 0018h). Once in the Interrupt Service Routine, the source(s) of the interrupt can be determined by polling the interrupt flag bits. The interrupt flag bits must be cleared in software before re-enabling interrupts to avoid recursive interrupts.

The "return from interrupt" instruction, RETFIE, exits the interrupt routine and sets the GIE bit (GIEH or GIEL if priority levels are used) which re-enables interrupts.

For external interrupt events, such as the INT pins or the PORTB input change interrupt, the interrupt latency will be three to four instruction cycles. The exact latency is the same for one or two-cycle instructions. Individual interrupt flag bits are set regardless of the status of their corresponding enable bit or the GIE bit.

Note: Do not use the MOVFF instruction to modify any of the Interrupt Control registers while any interrupt is enabled. Doing so may cause erratic microcontroller behavior.





8.1 INTCON Registers

The INTCON registers are readable and writable registers which contain various enable, priority and flag bits.

Note: Interrupt flag bits are set when an interrupt condition occurs regardless of the state of its corresponding enable bit or the global interrupt enable bit. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt. This feature allows for software polling.

REGISTER 8-1: INTCON: INTERRUPT CONTROL REGISTER

GIE/GIEH PEIE/GIEL TMROIE INTOIE RBIE TMROIF II	NTOIF RBIF ⁽¹⁾
	NTOIF RBIF ⁽¹⁾
bit 7	bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	GIE/GIEH: Global Interrupt Enable bit <u>When IPEN = 0:</u> 1 = Enables all unmasked interrupts 0 = Disables all interrupts <u>When IPEN = 1:</u> 1 = Enables all high priority interrupts 0 = Disables all high priority interrupts
L H 0	0 = Disables all interrupts
bit 6	PEIE/GIEL: Peripheral Interrupt Enable bit When IPEN = 0:
	1 = Enables all unmasked peripheral interrupts 0 = Disables all peripheral interrupts
	When IPEN = 1:
	 1 = Enables all low priority peripheral interrupts 0 = Disables all low priority peripheral interrupts
bit 5	TMR0IE: TMR0 Overflow Interrupt Enable bit
	1 = Enables the TMR0 overflow interrupt0 = Disables the TMR0 overflow interrupt
bit 4	INTOIE: INTO External Interrupt Enable bit
	 1 = Enables the INT0 external interrupt 0 = Disables the INT0 external interrupt
bit 3	RBIE: RB Port Change Interrupt Enable bit
	1 = Enables the RB port change interrupt
	0 = Disables the RB port change interrupt
bit 2	TMR0IF: TMR0 Overflow Interrupt Flag bit
	 1 = TMR0 register has overflowed (must be cleared in software) 0 = TMR0 register did not overflow
bit 1	INTOIF: INTO External Interrupt Flag bit
	 1 = The INT0 external interrupt occurred (must be cleared in software) 0 = The INT0 external interrupt did not occur
bit 0	RBIF: RB Port Change Interrupt Flag bit ⁽¹⁾
	 1 = At least one of the RB7:RB4 pins changed state (must be cleared in software) 0 = None of the RB7:RB4 pins have changed state
Note 1:	A mismatch condition will continue to set this bit. Reading PORTB will end the mismatch condition and

Note 1: A mismatch condition will continue to set this bit. Reading PORTB will end the mismatch condition and allow the bit to be cleared.

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1			
RBPU	INTEDG0	INTEDG1	INTEDG2	INTEDG3	TMR0IP	INT3IP	RBIP			
bit 7							bit			
Legend:										
R = Readabl	e bit	W = Writable	bit	U = Unimplem	nented bit, rea	d as '0'				
R = Readable bit -n = Value at POR		'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	own			
bit 7	RBPU: PORT	B Pull-up Enal	ole bit							
		B pull-ups are								
	0 = PORTB p	oull-ups are ena	abled by individ	dual port latch v	alues					
bit 6	INTEDG0: Ex	ternal Interrupt	0 Edge Select	t bit						
		on rising edge								
	0 = Interrupt on falling edge									
bit 5	INTEDG1: External Interrupt 1 Edge Select bit									
	 Interrupt on rising edge Interrupt on falling edge 									
bit 4										
DIL 4	INTEDG2: External Interrupt 2 Edge Select bit 1 = Interrupt on rising edge									
	0 = Interrupt on falling edge									
bit 3	•	ternal Interrupt	3 Edge Select	t bit						
	1 = Interrupt on rising edge									
	0 = Interrupt	0 = Interrupt on falling edge								
bit 2	TMR0IP: TM	R0 Overflow Int	errupt Priority	bit						
	1 = High priority									
	0 = Low prior	•								
bit 1		External Interr	upt Priority bit							
	1 = High priority									
L 1 0	0 = Low priority									
bit 0	RBIP: RB Port Change Interrupt Priority bit									
	1 = High prio 0 = Low prior									
		•••								

REGISTER 8-2: INTCON2: INTERRUPT CONTROL REGISTER 2

enable bit or the global interrupt enable bit. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt. This feature allows for software polling.

REGISTER 8-3: INTCON3: INTERRUPT CONTROL REGISTER 3

R/W-1	R/W-1	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
INT2IP	INT1IP	INT3IE	INT2IE	INT1IE	INT3IF	INT2IF	INT1IF
bit 7							bit (
Legend:							
R = Reada	able bit	W = Writable	bit	U = Unimplen	nented bit, rea	d as '0'	
-n = Value	at POR	'1' = Bit is set		'0' = Bit is clea		x = Bit is unkr	nown
bit 7	INT2IP: INT2	External Interr	upt Priority bit				
	1 = High pric 0 = Low prio	•					
bit 6	INT1IP: INT1	External Interr	upt Priority bit				
	1 = High pric 0 = Low prio	,					
bit 5	INT3IE: INT3	External Interr	upt Enable bit				
		the INT3 extern the INT3 extern					
bit 4	INT2IE: INT2	External Interr	upt Enable bit				
		the INT2 extern					
bit 3		the INT2 exter External Interr					
Sit 0	1 = Enables	the INT1 extern	nal interrupt				
h it 0		the INT1 exter	•				
bit 2		External Interr		must be cleared	h in coftwara)		
		3 external interi			a in Soltware)		
bit 1	INT2IF: INT2	External Interr	upt Flag bit				
		2 external inter 2 external inter	•	must be cleared cur	d in software)		
bit 0	INT1IF: INT1	External Interr	upt Flag bit				
		1 external intern 1 external intern		must be cleared cur	d in software)		
Note:	Interrupt flag bits	s are set when	an interrupt or	andition occurs	regardless of	the state of its	corresponding
1010	enable bit or the are clear prior to	global interrupt	enable bit. Us	er software sho	ould ensure the	e appropriate int	

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8.2 PIR Registers

The PIR registers contain the individual flag bits for the peripheral interrupts. Due to the number of peripheral interrupt sources, there are three Peripheral Interrupt Request (Flag) registers (PIR1, PIR2, PIR3).

- Note 1: Interrupt flag bits are set when an interrupt condition occurs regardless of the state of its corresponding enable bit or the Global Interrupt Enable bit, GIE (INTCON<7>).
 - 2: User software should ensure the appropriate interrupt flag bits are cleared prior to enabling an interrupt and after servicing that interrupt.

REGISTER 8-4: PIR1: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 1

U-0	R/W-0	R-0	R-0	R/W-0	U-0	R/W-0	R/W-0
_	ADIF	RC1IF	TX1IF	SSPIF	—	TMR2IF	TMR1IF
bit 7							bit 0

Legend:				
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown	

bit 7	Unimplemented: Read as '0'
bit 6	ADIF: A/D Converter Interrupt Flag bit
	 1 = An A/D conversion completed (must be cleared in software) 0 = The A/D conversion is not complete
bit 5	RC1IF: EUSART Receive Interrupt Flag bit
	 1 = The EUSART receive buffer, RCREG1, is full (cleared when RCREG1 is read) 0 = The EUSART receive buffer is empty
bit 4	TX1IF: EUSART Transmit Interrupt Flag bit
	1 = The EUSART transmit buffer, TXREG1, is empty (cleared when TXREG1 is written)0 = The EUSART transmit buffer is full
bit 3	SSPIF: Master Synchronous Serial Port Interrupt Flag bit
	1 = The transmission/reception is complete (must be cleared in software)0 = Waiting to transmit/receive
bit 2	Unimplemented: Read as '0'
bit 1	TMR2IF: TMR2 to PR2 Match Interrupt Flag bit
	1 = TMR2 to PR2 match occurred (must be cleared in software)0 = No TMR2 to PR2 match occurred
bit 0	TMR1IF: TMR1 Overflow Interrupt Flag bit
	1 = TMR1 register overflowed (must be cleared in software)0 = TMR1 register did not overflow

REGISTER 8-5: PIR2: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 2

R/W-0	R/W-0	U-0	U-0	R/W-0	R/W-0	R/W-0	U-0
OSCFIF	CMIF	—	—	BCLIF	LVDIF	TMR3IF	—
bit 7							bit 0

Legend:				
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown	

bit 7	OSCFIF: Oscillator Fail Interrupt Flag bit 1 = Device oscillator failed, clock input has changed to INTOSC (must be cleared in software) 0 = Device clock operating
bit 6	CMIF: Comparator Interrupt Flag bit 1 = Comparator input has changed (must be cleared in software)
	0 = Comparator input has not changed
bit 5-4	Unimplemented: Read as '0'
bit 3	BCLIF: Bus Collision Interrupt Flag bit
	 1 = A bus collision occurred (must be cleared in software) 0 = No bus collision occurred
bit 2	LVDIF: Low-Voltage Detect Interrupt Flag bit
	 1 = A low-voltage condition occurred (must be cleared in software) 0 = The device voltage is above the regulator's low-voltage trip point
bit 1	TMR3IF: TMR3 Overflow Interrupt Flag bit
	1 = TMR3 register overflowed (must be cleared in software)
	0 = TMR3 register did not overflow
bit 0	Unimplemented: Read as '0'

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U-0	R/W-0	R-0	R-0	U-0	R/W-0	R/W-0	U-0			
_	LCDIF	RC2IF	TX2IF	_	CCP2IF	CCP1IF	_			
oit 7							bi			
Legend:										
R = Reada	ble bit	W = Writable	bit	U = Unimple	mented bit, rea	d as '0'				
R = Readable bit -n = Value at POR		'1' = Bit is set		'0' = Bit is cle	eared	x = Bit is unkno	own			
oit 7	•	ted: Read as '								
oit 6						tatic mode is sele	ected)			
				t be cleared in	software)					
		a of all COMs is	• •							
bit 5		ART Receive In				00: 1				
		SART receive bi		2, is full (cleare	ed when RCRE	G2 is read)				
oit 4	TX2IF: AUSART Transmit Interrupt Flag bit									
	1 = The AUSART transmit buffer, TXREG2, is empty (cleared when TXREG2 is written)									
	0 = The AUS	ART transmit b	ouffer is full							
bit 3	Unimplemen	ted: Read as '	כ'							
oit 2	CCP2IF: CCF	P2 Interrupt Fla	g bit							
	Capture mode:									
	1 = A TMR1/TMR3 register capture occurred (must be cleared in software)									
	 0 = No TMR1/TMR3 register capture occurred Compare mode: 									
	<u>Compare mode:</u> 1 = A TMR1/TMR3 register compare match occurred (must be cleared in software)									
	0 = No TMR1/TMR3 register compare match occurred									
	<u>PWM mode:</u>									
	Unused in thi	s mode.								
bit 1	CCP1IF: CCP1 Interrupt Flag bit									
	Capture mode:									
	1 = A TMR1/TMR3 register capture occurred (must be cleared in software)									
	 0 = No TMR1/TMR3 register capture occurred Compare mode: 									
			compare mat	ch occurred (m	ust be cleared	in software)				
		1/TMR3 registe				······································				
	PWM mode:									
	Unused in thi	s mode.								
bit 0	Unimplemen	ted: Read as '	יר							

REGISTER 8-6: PIR3: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 3

8.3 PIE Registers

The PIE registers contain the individual enable bits for the peripheral interrupts. Due to the number of peripheral interrupt sources, there are three Peripheral Interrupt Enable registers (PIE1, PIE2, PIE3). When IPEN = 0, the PEIE bit must be set to enable any of these peripheral interrupts.

REGISTER 8-7: PIE1: PERIPHERAL INTERRUPT ENABLE REGISTER 1

U-0	R/W-0	R/W-0	R/W-0	R/W-0	U-0	R/W-0	R/W-0
—	ADIE	RC1IE	TX1IE	SSPIE	—	TMR2IE	TMR1IE
bit 7							bit 0

Legend:			
R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'			
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	Unimplemented: Read as '0'
bit 6	ADIE: A/D Converter Interrupt Enable bit
	1 = Enables the A/D interrupt
	0 = Disables the A/D interrupt
bit 5	RC1IE: EUSART Receive Interrupt Enable bit
	1 = Enables the EUSART receive interrupt
	0 = Disables the EUSART receive interrupt
bit 4	TX1IE: EUSART Transmit Interrupt Enable bit
	1 = Enables the EUSART transmit interrupt
	0 = Disables the EUSART transmit interrupt
bit 3	SSPIE: Master Synchronous Serial Port Interrupt Enable bit
	1 = Enables the MSSP interrupt
	0 = Disables the MSSP interrupt
bit 2	Unimplemented: Read as '0'
bit 1	TMR2IE: TMR2 to PR2 Match Interrupt Enable bit
	1 = Enables the TMR2 to PR2 match interrupt
	0 = Disables the TMR2 to PR2 match interrupt
bit 0	TMR1IE: TMR1 Overflow Interrupt Enable bit
	1 = Enables the TMR1 overflow interrupt
	0 = Disables the TMR1 overflow interrupt

R/W-0	R/W-0	U-0	U-0	R/W-0	R/W-0	R/W-0	U-0
OSCFIE	CMIE	—	—	BCLIE	LVDIE	TMR3IE	—
bit 7							bit 0
Legend:							
R = Readable	e bit	W = Writable	bit	U = Unimplem	nented bit, read	d as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkn	own
bit 7	OSCFIE: Osc	illator Fail Inter	rupt Enable bit	t			
	1 = Enabled						
	0 = Disabled						
bit 6	•	arator Interrupt	Enable bit				
	1 = Enabled						
	0 = Disabled	ted. Dood oo '	.,				
bit 5-4	•	ted: Read as 'o					
bit 3	BCLIE: Bus Collision Interrupt Enable bit						
	1 = Enabled 0 = Disabled						
bit 2		/oltago Dotoot I	ntorrunt Enchl	o hit			
Dit 2	1 = Enabled	/oltage Detect I	nterrupt Enabl	e bit			
	0 = Disabled						
bit 1	TMR3IE: TMF	R3 Overflow Int	errupt Enable I	bit			
	1 = Enabled		·				
	0 = Disabled						
bit 0	Unimplemen	ted: Read as 'o)'				

REGISTER 8-8: PIE2: PERIPHERAL INTERRUPT ENABLE REGISTER 2

REGISTER 8-9: PIE3: PERIPHERAL INTERRUPT ENABLE REGISTER 3

U-0	R/W-0	R-0	R-0	U-0	R/W-0	R/W-0	U-0			
_	LCDIE	RC2IE	TX2IE	—	CCP2IE	CCP1IE				
oit 7							bit 0			
Legend:			.,			(0)				
R = Readabl		W = Writable b	Dit	•	nented bit, read					
n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkno	own			
·. -										
oit 7	•	nted: Read as 'c								
oit 6		Interrupt Enable	bit (valid whe	n Type-B wavefo	orm with Non-S	tatic mode is sel	ected)			
	1 = Enabled	-								
	0 = Disable	d								
oit 5	RC2IE: AUS	RC2IE: AUSART Receive Interrupt Enable bit								
	1 = Enabled									
	0 = Disable	d								
oit 4	TX2IE: AUS	ART Transmit In	terrupt Enable	bit						
	1 = Enabled									
	0 = Disable	d								
oit 3	Unimpleme	nted: Read as 'o	,							
oit 2	CCP2IE: CC	P2 Interrupt Ena	uble bit							
	1 = Enabled	ł								
	0 = Disable	d								
oit 1	CCP1IE: CC	P1 Interrupt Ena	uble bit							
		the CCP1 interru								
		the CCP1 interr								

8.4 IPR Registers

The IPR registers contain the individual priority bits for the peripheral interrupts. Due to the number of peripheral interrupt sources, there are three Peripheral Interrupt Priority registers (IPR1, IPR2, IPR3). Using the priority bits requires that the Interrupt Priority Enable (IPEN) bit be set.

REGISTER 8-10: IPR1: PERIPHERAL INTERRUPT PRIORITY REGISTER 1

U-0	R/W-1	R/W-1	R/W-1	R/W-1	U-0	R/W-1	R/W-1			
_	ADIP	RC1IP	TX1IP	SSPIP	—	TMR2IP	TMR1IP			
bit 7						•	bit (
Legend:										
R = Readable	e bit	W = Writable	bit	U = Unimplen	nented bit, rea	ıd as '0'				
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown			
bit 7	Unimplemen	ted: Read as '	כ'							
bit 6	ADIP: A/D Co	onverter Interru	pt Priority bit							
	1 = High prio	ority								
	0 = Low prior	rity								
bit 5	RC1IP: EUSART Receive Interrupt Priority bit									
	1 = High priority									
	0 = Low prior	•								
bit 4	TX1IP: EUSA	RT Transmit Ir	terrupt Priority	/ bit						
	1 = High prio	•								
	0 = Low prior									
bit 3		-	Serial Port In	terrupt Priority I	bit					
	1 = High prio									
1.11.0	0 = Low prior	-								
bit 2	•	ted: Read as '								
bit 1	TMR2IP: TMR2 to PR2 Match Interrupt Priority bit									
	1 = High priority									
hit O	0 = Low prior	•	orrupt Drigrith	h.it						
bit 0		R1 Overflow Int	errupt Priority	DIL						
	1 = High prio 0 = Low prior	•								

REGISTER 8-11: IPR2: PERIPHERAL INTERRUPT PRIORITY REGISTER 2

HVW-1 D-0 D-0 HVW-1 HV bit 7 DSCEIP: Oscillator Fail Interrupt Priority bit 1 = High priority 0 = Low		D/M/ 1		11.0	D/4/ 1	D/4/ 1	D () () 1					
bit 7 Legend: R = Readable bit W = Writable bit U = Unimplemented bit, read as '0' -n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown bit 7 OSCFIP: Oscillator Fail Interrupt Priority bit 1 = High priority 0 = Low priority bit 6 CMIP: Comparator Interrupt Priority bit 1 = High priority 0 = Low priority bit 5-4 Unimplemented: Read as '0' bit 3 BCLIP: Bus Collision Interrupt Priority bit bit 3 BCLIP: Bus Collision Interrupt Priority bit 1 = High priority 0 = Low priority 0 = Low priority bit 2 LVDIP: Low-Voltage Detect Interrupt Priority bit 1 = High priority 0 = Low priority bit 1 TMR3IP: TMR3 Overflow Interrupt Priority bit 1 = High priority 0 = Low priority	R/W-1	R/W-1	U-0	U-0	R/W-1	R/W-1	R/W-1	U-0				
Legend: R = Readable bit W = Writable bit U = Unimplemented bit, read as '0' -n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown bit 7 OSCFIP: Oscillator Fail Interrupt Priority bit 1 = High priority 0 = Low priority bit 6 CMIP: Comparator Interrupt Priority bit 1 = High priority 0 = Low priority bit 5-4 Unimplemented: Read as '0' bit 3 BCLIP: Bus Collision Interrupt Priority bit bit 3 BCLIP: Bus Collision Interrupt Priority bit 1 = High priority 0 = Low priority 0 = Low priority bit 2 LVDIP: Low-Voltage Detect Interrupt Priority bit 1 = High priority 0 = Low priority bit 1 TMR3IP: TMR3 Overflow Interrupt Priority bit 1 = High priority 0 = Low priority		CMIP			BCLIP	LVDIP	TMR3IP					
R = Readable bit W = Writable bit U = Unimplemented bit, read as '0' -n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown bit 7 OSCFIP: Oscillator Fail Interrupt Priority bit 1 = High priority 0 = Bit is cleared x = Bit is unknown bit 7 OSCFIP: Oscillator Fail Interrupt Priority bit 1 = High priority 0 = Low priority bit 6 CMIP: Comparator Interrupt Priority bit 1 = High priority 0 = Low priority bit 5-4 Unimplemented: Read as '0' bit 3 BCLIP: Bus Collision Interrupt Priority bit 1 = High priority bit 3 BCLIP: Low-Voltage Detect Interrupt Priority bit 1 = High priority 0 = Low priority bit 4 LVDIP: Low-Voltage Detect Interrupt Priority bit 1 = High priority 0 = Low priority bit 1 TMR3IP: TMR3 Overflow Interrupt Priority bit 1 = High priority 0 = Low priority bit 1 TIMR3IP: TMR3 Overflow Interrupt Priority bit 1 = High priority 0 = Low priority	bit 7							bit 0				
R = Readable bit W = Writable bit U = Unimplemented bit, read as '0' -n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown bit 7 OSCFIP: Oscillator Fail Interrupt Priority bit 1 = High priority 0 = Low priority bit 6 CMIP: Comparator Interrupt Priority bit 1 = High priority 0 = Low priority bit 5 CMIP: Comparator Interrupt Priority bit 1 = High priority 0 = Low priority bit 5-4 Unimplemented: Read as '0' bit 3 BCLIP: Bus Collision Interrupt Priority bit 1 = High priority 0 = Low priority bit 1 = High priority 0 = Low priority 0 = Low priority bit 1 = High priority 0 = Low priority 0 = Low priority bit 1 = High priority 0 = Low priority 0 = Low priority bit 1 = High priority 0 = Low priority 0 = Low priority bit 1 = High priority 0 = Low priority 0 = Low priority bit 1 = High priority 0 = Low priority 0 = Low priority bit 1 = High priority												
-n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown bit 7 OSCFIP: Oscillator Fail Interrupt Priority bit 1 = High priority 0 = Low priority bit 6 CMIP: Comparator Interrupt Priority bit 1 = High priority 0 = Low priority bit 5 CMIP: Comparator Interrupt Priority bit 1 = High priority 0 = Low priority bit 5-4 Unimplemented: Read as '0' bit 3 BCLIP: Bus Collision Interrupt Priority bit 1 = High priority 0 = Low priority bit 1 = High priority bit 2 LVDIP: Low-Voltage Detect Interrupt Priority bit 1 = High priority 0 = Low priority 0 = Low priority bit 1 t = High priority 0 = Low priority bit 1 t = High priority 0 = Low priority bit 1 t = High priority 0 = Low priority bit 1 t = High priority 0 = Low priority bit 1 t = High priority 0 = Low priority bit 1	•											
bit 7 OSCFIP: Oscillator Fail Interrupt Priority bit 1 = High priority 0 = Low priority bit 6 CMIP: Comparator Interrupt Priority bit 1 = High priority 0 = Low priority bit 5-4 Unimplemented: Read as '0' bit 3 BCLIP: Bus Collision Interrupt Priority bit 1 = High priority 0 = Low priority bit 2 LVDIP: Low-Voltage Detect Interrupt Priority bit 1 = High priority 0 = Low priority bit 1 TMR3IP: TMR3 Overflow Interrupt Priority bit 1 = High priority 0 = Low priority bit 1 = High priority 0 = Low priority				bit	•							
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bit 6CMIP: Comparator Interrupt Priority bit1 = High priority0 = Low prioritybit 5-4Unimplemented: Read as '0'bit 3BCLIP: Bus Collision Interrupt Priority bit1 = High priority0 = Low prioritybit 2LVDIP: Low-Voltage Detect Interrupt Priority bit1 = High priority0 = Low prioritybit 1TMR3IP: TMR3 Overflow Interrupt Priority bit1 = High priority0 = Low prioritybit 1Difference0 = Low prioritybit 10 = Low priority												
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 bit 5-4 Unimplemented: Read as '0' bit 3 BCLIP: Bus Collision Interrupt Priority bit High priority Low priority bit 2 LVDIP: Low-Voltage Detect Interrupt Priority bit High priority Low priority bit 1 TMR3IP: TMR3 Overflow Interrupt Priority bit High priority Low priority 	bit 6	CMIP: Comp	CMIP: Comparator Interrupt Priority bit									
bit 5-4 Unimplemented: Read as '0' bit 3 BCLIP: Bus Collision Interrupt Priority bit 1 = High priority 0 = Low priority bit 2 LVDIP: Low-Voltage Detect Interrupt Priority bit 1 = High priority 0 = Low priority bit 1 TMR3IP: TMR3 Overflow Interrupt Priority bit 1 = High priority 0 = Low priority bit 1 Explanation of the priority bit 1 = High priority 0 = Low priority		• .	•									
bit 3 BCLIP: Bus Collision Interrupt Priority bit 1 = High priority 0 = Low priority bit 2 LVDIP: Low-Voltage Detect Interrupt Priority bit 1 = High priority 0 = Low priority bit 1 TMR3IP: TMR3 Overflow Interrupt Priority bit 1 = High priority 0 = Low priority		0 = Low prio	ority									
<pre>1 = High priority 0 = Low priority bit 2 LVDIP: Low-Voltage Detect Interrupt Priority bit 1 = High priority 0 = Low priority bit 1 TMR3IP: TMR3 Overflow Interrupt Priority bit 1 = High priority 0 = Low priority</pre>	bit 5-4	Unimplemer	nted: Read as '0	,								
 bit 2 LVDIP: Low-Voltage Detect Interrupt Priority bit 1 = High priority 0 = Low priority bit 1 TMR3IP: TMR3 Overflow Interrupt Priority bit 1 = High priority 0 = Low priority 	bit 3	BCLIP: Bus	Collision Interrup	ot Priority bit								
bit 2 LVDIP: Low-Voltage Detect Interrupt Priority bit 1 = High priority 0 = Low priority bit 1 TMR3IP: TMR3 Overflow Interrupt Priority bit 1 = High priority 0 = Low priority		1 = High prio	1 = High priority									
 1 = High priority 0 = Low priority bit 1 TMR3IP: TMR3 Overflow Interrupt Priority bit 1 = High priority 0 = Low priority 		0 = Low prior	ority									
 bit 1 TMR3IP: TMR3 Overflow Interrupt Priority bit 1 = High priority 0 = Low priority 	bit 2	LVDIP: Low-	LVDIP: Low-Voltage Detect Interrupt Priority bit									
bit 1 TMR3IP: TMR3 Overflow Interrupt Priority bit 1 = High priority 0 = Low priority		1 = High price										
 1 = High priority 0 = Low priority 		0 = Low prior	0 = Low priority									
0 = Low priority	bit 1	TMR3IP: TM	TMR3IP: TMR3 Overflow Interrupt Priority bit									
		1 = High price	ority									
bit 0 Unimplemented: Read as '0'		0 = Low prior	ority									
	bit 0	Unimplemer	nted: Read as '0	,								

U-0	R/W-0	R-0	R-0	U-0	R/W-1	R/W-1	U-0	
	LCDIP	RC2IP	TX2IP		CCP2IP	CCP1IP	—	
bit 7							bit 0	
Legend:								
R = Readab	R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'							
-n = Value a	it POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkn	iown	
bit 7	Unimplemen	ted: Read as '	כ'					
bit 6	LCDIP: LCD I	nterrupt Priority	v bit (valid wher	n Type-B wavef	orm with Non-S	Static mode is se	elected)	
	1 = High prio							
	0 = Low prior	rity						
bit 5		ART Receive P	riority Flag bit					
	1 = High prio	•						
	0 = Low prior	•						
bit 4		RT Transmit Ir	iterrupt Priority	bit				
	1 = High prio 0 = Low prior							
bit 3	-	ted: Read as '	<u>`</u>					
	-							
bit		P2 Interrupt Pri	only dit					
	1 = High prio 0 = Low prior	•						
bit	•	P1 Interrupt Pri	ority bit					
NIL.	1 = High prio	-						
	0 = Low prior	•						
bit 0	•	ted: Read as '	o'					

REGISTER 8-12: IPR3: PERIPHERAL INTERRUPT PRIORITY REGISTER 3

8.5 RCON Register

The RCON register contains bits used to determine the cause of the last Reset or wake-up from Idle or Sleep modes. RCON also contains the bit that enables interrupt priorities (IPEN).

REGISTER 8-13: RCON: RESET CONTROL REGISTER

R/W-0	U-0	U-0	R/W-1	R-1	R-1	R/W-0	R/W-0
IPEN	—	_	RI	TO	PD	POR	BOR
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	t, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	IPEN: Interrupt Priority Enable bit 1 = Enable priority levels on interrupts 0 = Disable priority levels on interrupts (PIC16CXXX Compatibility mode)
bit 6-5	Unimplemented: Read as '0'
bit 4	RI: RESET Instruction Flag bit
	For details of bit operation, see Register 4-1.
bit 3	TO: Watchdog Timer Time-out Flag bit
	For details of bit operation, see Register 4-1.
bit 2	PD: Power-Down Detection Flag bit
	For details of bit operation, see Register 4-1.
bit 1	POR: Power-on Reset Status bit
	For details of bit operation, see Register 4-1.
bit 0	BOR: Brown-out Reset Status bit
	For details of bit operation, see Register 4-1.

8.6 INTx Pin Interrupts

External interrupts on the RB0/INT0, RB1/INT1, RB2/INT2 and RB3/INT3 pins are edge-triggered. If the corresponding INTEDGx bit in the INTCON2 register is set (= 1), the interrupt is triggered by a rising edge; if the bit is clear, the trigger is on the falling edge. When a valid edge appears on the RBx/INTx pin, the corresponding flag bit, INTxIF, is set. This interrupt can be disabled by clearing the corresponding enable bit, INTxIE. Flag bit, INTxIF, must be cleared in software in the Interrupt Service Routine before re-enabling the interrupt.

All external interrupts (INT0, INT1, INT2 and INT3) can wake-up the processor from the power-managed modes if bit INTxIE was set prior to going into the power-managed modes. If the Global Interrupt Enable bit, GIE, is set, the processor will branch to the interrupt vector following wake-up.

Interrupt priority for INT1, INT2 and INT3 is determined by the value contained in the interrupt priority bits, INT1IP (INTCON3<6>), INT2IP (INTCON3<7>) and INT3IP (INTCON2<1>). There is no priority bit associated with INT0. It is always a high priority interrupt source.

8.7 TMR0 Interrupt

In 8-bit mode (which is the default), an overflow in the TMR0 register (FFh \rightarrow 00h) will set flag bit, TMR0IF. In 16-bit mode, an overflow in the TMR0H:TMR0L register pair (FFFFh \rightarrow 0000h) will set TMR0IF. The interrupt can be enabled/disabled by setting/clearing enable bit, TMR0IE (INTCON<5>). Interrupt priority for Timer0 is determined by the value contained in the interrupt priority bit, TMR0IP (INTCON2<2>). See Section 10.0 "Timer0 Module" for further details on the Timer0 module.

8.8 **PORTB Interrupt-on-Change**

An input change on PORTB<7:4> sets flag bit, RBIF (INTCON<0>). The interrupt can be enabled/disabled by setting/clearing enable bit, RBIE (INTCON<3>). Interrupt priority for PORTB interrupt-on-change is determined by the value contained in the interrupt priority bit, RBIP (INTCON2<0>).

8.9 Context Saving During Interrupts

During interrupts, the return PC address is saved on the stack. Additionally, the WREG, STATUS and BSR registers are saved on the Fast Return Stack. If a fast return from interrupt is not used (see **Section 5.3 "Data Memory Organization"**), the user may need to save the WREG, STATUS and BSR registers on entry to the Interrupt Service Routine. Depending on the user's application, other registers may also need to be saved. Example 8-1 saves and restores the WREG, STATUS and BSR registers during an Interrupt Service Routine.

	EXAMPLE 8-1:	SAVING STATUS, WREG AND BSR REGISTERS IN RAM
--	--------------	--

MOVWF MOVFF MOVFF	W_TEMP STATUS, STATUS_TEMP BSR, BSR_TEMP	; W_TEMP is in virtual bank ; STATUS_TEMP located anywhere ; BSR_TMEP located anywhere
; ; USER :	ISR CODE	
, MOVFF MOVF MOVFF	BSR_TEMP, BSR W_TEMP, W STATUS_TEMP, STATUS	; Restore BSR ; Restore WREG ; Restore STATUS

9.0 I/O PORTS

Depending on the device selected and features enabled, there are up to nine ports available. Some pins of the I/O ports are multiplexed with an alternate function from the peripheral features on the device. In general, when a peripheral is enabled, that pin may not be used as a general purpose I/O pin.

Each port has three memory mapped registers for its operation:

- TRIS register (Data Direction register)
- PORT register (reads the levels on the pins of the device)
- LAT register (Output Latch register)

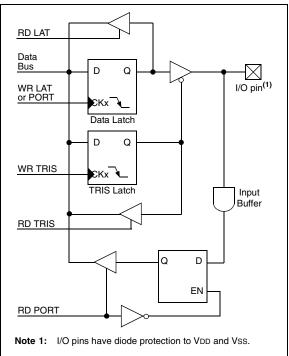
Reading the PORT register reads the current status of the pins, whereas writing to the PORT register writes to the Output Latch (LAT) register.

Setting a TRIS bit (= 1) makes the corresponding PORT pin an input (i.e., put the corresponding output driver in a high-impedance mode). Clearing a TRIS bit (= 0) makes the corresponding PORT pin an output (i.e., put the contents of the corresponding LAT bit on the selected pin).

The Data Latch (LAT register) is useful for read-modify-write operations on the value that the I/O pins are driving. Read-modify-write operations on the LAT register read and write the latched output value for the PORT register.

A simplified model of a generic I/O port, without the interfaces to other peripherals, is shown in Figure 9-1.

FIGURE 9-1: GENERIC I/O PORT OPERATION



9.1 I/O Port Pin Capabilities

When developing an application, the capabilities of the port pins must be considered. Outputs on some pins have higher output drive strength than others. Similarly, some pins can tolerate higher than VDD input levels.

9.1.1 INPUT PINS AND VOLTAGE CONSIDERATIONS

The voltage tolerance of pins used as device inputs is dependent on the pin's input function. Pins that are used as digital only inputs are able to handle DC voltages up to 5.5V, a level typical for digital logic circuits. In contrast, pins that also have analog input functions of any kind can only tolerate voltages up to VDD. Voltage excursions beyond VDD on these pins should be avoided.

Table 9-1 summarizes the input voltage capabilities. Refer to **Section 25.0 "Electrical Characteristics**" for more details.

TABLE 9-1: INPUT VOLTAGE TOLERANCE

PORT or Pin	Tolerated Input	Description
PORTA<7:0>	Vdd	Only VDD input levels
PORTC<1:0>		tolerated.
PORTF<7:1>		
PORTB<7:0>	5.5V	Tolerates input levels
PORTC<7:2>		above VDD, useful for
PORTD<7:0>		most standard logic.
PORTE<7:0>		
PORTG<4:0>		
PORTH<7:0>(1)		
PORTJ<7:0> ⁽¹⁾		

Note 1: Not available on 64-pin devices.

9.1.2 PIN OUTPUT DRIVE

When used as digital I/O, the output pin drive strengths vary for groups of pins intended to meet the needs for a variety of applications. In general, there are three classes of output pins in terms of drive capability.

PORTB and PORTC, as well as PORTA<7:6>, are designed to drive higher current loads, such as LEDs. PORTD, PORTE and PORTJ can also drive LEDs but only those with smaller current requirements. PORTF, PORTG and PORTH, along with PORTA<5:0>, have the lowest drive level but are capable of driving normal digital circuit loads with a high input impedance. Regardless of which port it is located on, all output pins in LCD Segment or Common mode have sufficient output to directly drive a display.

Table 9-2 summarizes the output capabilities of the ports. Refer to the "Absolute Maximum Ratings" in Section 25.0 "Electrical Characteristics" for more details.

TABLE 9-2:OUTPUT DRIVE LEVELS FOR
VARIOUS PORTS

Low	Medium	High
PORTA<5:0>	PORTD	PORTA<7:6>
PORTF	PORTE	PORTB
PORTG	PORTJ ⁽¹⁾	PORTC
PORTH ⁽¹⁾		

Note 1: Not available on 64-pin devices.

9.1.3 PULL-UP CONFIGURATION

Four of the I/O ports (PORTB, PORTD, PORTE and PORTJ) implement configurable weak pull-ups on all pins. These are internal pull-ups that allow floating digital input signals to be pulled to a consistent level without the use of external resistors.

The pull-ups are enabled with a single bit for each of the ports: RBPU (INTCON2<7>) for PORTB, and RDPU, REPU and PJPU (PORTG<7:5>) for the other ports.

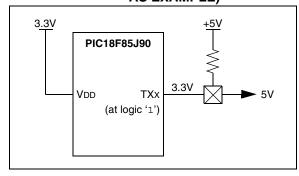
9.1.4 OPEN-DRAIN OUTPUTS

The output pins for several peripherals are also equipped with a configurable, open-drain output option. This allows the peripherals to communicate with external digital logic, operating at a higher voltage level, without the use of level translators.

The open-drain option is implemented on port pins specifically associated with the data and clock outputs of the USARTs, the MSSP module (in SPI mode) and the CCP modules. This option is selectively enabled by setting the open-drain control bit for the corresponding module in TRISG and LATG. Their configuration is discussed in more detail in Section 9.4 "PORTC, TRISC and LATC Registers", Section 9.6 "PORTE, TRISE and LATE Registers" and Section 9.8 "PORTG, TRISG and LATG Registers".

When the open-drain option is required, the output pin must also be tied through an external pull-up resistor provided by the user to a higher voltage level, up to 5V (Figure 9-2). When a digital logic high signal is output, it is pulled up to the higher voltage level.

FIGURE 9-2: USING THE OPEN-DRAIN OUTPUT (USART SHOWN AS EXAMPLE)



9.2 PORTA, TRISA and LATA Registers

PORTA is an 8-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISA and LATA.

RA4/T0CKI is a Schmitt Trigger input. All other PORTA pins have TTL input levels and full CMOS output drivers.

The RA4 pin is multiplexed with the Timer0 clock input and one of the LCD segment drives. RA5 and RA3:RA0 are multiplexed with analog inputs for the A/D converter.

The operation of the analog inputs as A/D converter inputs is selected by clearing or setting the PCFG3:PCFG0 control bits in the ADCON1 register. The corresponding TRISA bits control the direction of these pins, even when they are being used as analog inputs. The user must ensure the bits in the TRISA register are maintained set when using them as analog inputs.

Note:	RA5 and RA3:RA0 are configured as							
	analog inputs on any Reset and are read							
	as '0'. RA4 is configured as a digital input.							

OSC2/CLKO/RA6 and OSC1/CLKI/RA7 normally serve as the external circuit connections for the external (primary) oscillator circuit (HS Oscillator modes), or the external clock input and output (EC Oscillator modes). In these cases, RA6 and RA7 are not available as digital I/O and their corresponding TRIS and LAT bits are read as '0'. When the device is configured to use INTOSC or INTRC as the default oscillator mode (FOSC2 Configuration bit is '0'), RA6 and RA7 are automatically configured as digital I/O; the oscillator and clock in/clock out functions are disabled.

RA1, RA4 and RA5 are multiplexed with LCD segment drives, controlled by bits in the LCDSE1 and LCDSE2 registers. I/O port functionality is only available when the LCD segments are disabled.

EXAMPLE 9-1: INITIALIZING PORTA

CLRF	PORTA	; Initialize PORTA by
		; clearing output latches
CLRF	LATA	; Alternate method to
		; clear output data latches
MOVLW	07h	; Configure A/D
MOVWF	ADCON1	; for digital inputs
MOVLW	0BFh	; Value used to initialize
		; data direction
MOVWF	TRISA	; Set RA<7, 5:0> as inputs,
		; RA<6> as output
		-

PIC18F85J90 FAMILY

Pin Name	Function	TRIS Setting	I/O	I/O Type	Description
RA0/AN0	RA0	0	0	DIG	LATA<0> data output; not affected by analog input.
		1	I	TTL	PORTA<0> data input; disabled when analog input enabled.
	AN0	1	I	ANA	A/D input channel 0. Default input configuration on POR; does not affect digital output.
RA1/AN1/SEG18	RA1	0	0	DIG	LATA<1> data output; not affected by analog input.
		1	I	TTL	PORTA<1> data input; disabled when analog input enabled.
	AN1	1	I	ANA	A/D input channel 1. Default input configuration on POR; does not affect digital output.
	SEG18	x	0	ANA	LCD segment 18 output; disables all other pin functions.
RA2/AN2/VREF-	RA2	0	0	DIG	LATA<2> data output; not affected by analog input.
RA2/AN2/VREF- RA3/AN3/VREF+ RA4/T0CKI/ SEG14		1	Ι	TTL	PORTA<2> data input; disabled when analog functions enabled.
	AN2	1	Ι	ANA	A/D input channel 2. Default input configuration on POR.
	VREF-	1	-	ANA	A/D and Comparator low reference voltage input.
RA3/AN3/VREF+	RA3	0	0	DIG	LATA<3> data output; not affected by analog input.
		1	Ι	TTL	PORTA<3> data input; disabled when analog input enabled.
	AN3	1	-	ANA	A/D input channel 3. Default input configuration on POR.
	VREF+	1	-	ANA	A/D and Comparator high reference voltage input.
RA4/T0CKI/	RA4	0	0	DIG	LATA<4> data output.
SEG14		1	-	ST	PORTA<4> data input. Default configuration on POR.
	TOCKI	x	-	ST	Timer0 clock input.
	SEG14	x	0	ANA	LCD segment 14 output; disables all other pin functions.
RA5/AN4/SEG15	RA5	0	0	DIG	LATA<5> data output; not affected by analog input.
RA3/AN3/VREF+ RA4/T0CKI/ SEG14		1	-	TTL	PORTA<5> data input; disabled when analog input enabled.
	AN4	1	-	ANA	A/D input channel 4. Default configuration on POR.
	SEG15	x	0	ANA	LCD segment 15 output; disables all other pin functions.
OSC2/CLKO/RA6	OSC2	x	0	ANA	Main oscillator feedback output connection (HS and HSPLL modes).
	CLKO	x	0	DIG	System cycle clock output (FOSC/4) (EC and ECPLL modes).
	RA6	0	0	DIG	LATA<6> data output; disabled when FOSC2 Configuration bit is set.
		1	-	TTL	PORTA<6> data input; disabled when FOSC2 Configuration bit is set
OSC1/CLKI/RA7	OSC1	x	I	ANA	Main oscillator input connection (HS and HSPLL modes).
	CLKI	х	Ι	ANA	Main external clock source input (EC and ECPLL modes).
	RA7	0	0	DIG	LATA<7> data output; disabled when FOSC2 Configuration bit is set.
		1	I	TTL	PORTA<7> data input; disabled when FOSC2 Configuration bit is set

TABLE 9-3: PORTA FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, TTL = TTL Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
PORTA	RA7 ⁽¹⁾	RA6 ⁽¹⁾	RA5	RA4	RA3	RA2	RA1	RA0	55
LATA	LATA7 ⁽¹⁾	LATA6 ⁽¹⁾	LATA5	LATA4	LATA3	LATA2	LATA1	LATA0	54
TRISA	TRISA7 ⁽¹⁾	TRISA6 ⁽¹⁾	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	54
ADCON1	—	_	VCFG1	VCFG0	PCFG3	PCFG2	PCFG1	PCFG0	53
LCDSE1	SE15	SE14	SE13	SE12	SE11	SE10	SE09	SE08	53
LCDSE2	SE23	SE22	SE21	SE20	SE19	SE18	SE17	SE16	53

Legend: — = unimplemented, read as '0'. Shaded cells are not used by PORTA.

Note 1: These bits are enabled depending on the oscillator mode selected. When not enabled as PORTA pins, they are disabled and read as 'X'.

9.3 PORTB, TRISB and LATB Registers

PORTB is an 8-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISB and LATB. All pins on PORTB are digital only and tolerate voltages up to 5.5V.

EXAMPLE 9-2: INITIALIZING PORTB

CLRF	PORTB	; Initialize PORTB by ; clearing output
CLRF	LATB	; data latches ; Alternate method ; to clear output : data latches
MOVLW	0CFh	; Value used to ; initialize data : direction
MOVWF	TRISB	; Set RB<3:0> as inputs ; RB<5:4> as outputs ; RB<7:6> as inputs

Each of the PORTB pins has a weak internal pull-up. A single control bit can turn <u>on all</u> the pull-ups. This is performed by clearing bit RBPU (INTCON2<7>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on a Power-on Reset.

Four of the PORTB pins (RB7:RB4) have an interrupt-on-change feature. Only pins configured as inputs can cause this interrupt to occur (i.e., any RB7:RB4 pin configured as an output is excluded from the interrupt-on-change comparison). The input pins (of RB7:RB4) are compared with the old value latched on the last read of PORTB. The "mismatch" outputs of RB7:RB4 are ORed together to generate the RB Port Change Interrupt with Flag bit, RBIF (INTCON<0>).

This interrupt can wake the device from power-managed modes. The user, in the Interrupt Service Routine, can clear the interrupt in the following manner:

- a) Any read or write of PORTB (except with the MOVFF (ANY), PORTB instruction). This will end the mismatch condition.
- b) Clear flag bit, RBIF.

A mismatch condition will continue to set flag bit, RBIF. Reading PORTB will end the mismatch condition and allow flag bit, RBIF, to be cleared.

The interrupt-on-change feature is recommended for wake-up on key depression operation and operations where PORTB is only used for the interrupt-on-change feature. Polling of PORTB is not recommended while using the interrupt-on-change feature.

RB5:RB0 are also multiplexed with LCD segment drives, controlled by bits in the LCDSE1 and LCDSE3 registers. I/O port functionality is only available when the LCD segments are disabled.

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Pin Name	Function	TRIS Setting	I/O	I/O Type	Description		
RB0/INT0/SEG30	RB0	0	0	DIG	LATB<0> data output.		
		1	Ι	TTL	PORTB<0> data input; weak pull-up when RBPU bit is cleared.		
	INT0	1	Ι	ST	External interrupt 0 input.		
	SEG30	x	0	DIG LATB<1> data output.			
RB1/INT1/SEG8	RB1	0	0	DIG	LATB<1> data output.		
		1	Ι	TTL	PORTB<1> data input; weak pull-up when RBPU bit is cleared.		
	INT1	1	I	ST	External interrupt 1 input.		
SEG8		x	0	ANA	LCD segment 8 output; disables all other pin functions.		
			LATB<2> data output.				
		1	Ι	TTL	PORTB<2> data input; weak pull-up when RBPU bit is cleared.		
	INT2	1	Ι	ST	External interrupt 2 input.		
	SEG9	x	0	ANA	LCD segment 9 output; disables all other pin functions.		
RB3/INT3/SEG10	RB3	0	0	DIG	LATB<3> data output.		
		1	Ι	TTL	PORTB<3> data input; weak pull-up when RBPU bit is cleared.		
	INT3	1	I	ST	External interrupt 3 input.		
	SEG10	x	0	ANA	LCD segment 10 output; disables all other pin functions.		
RB4/KBI0/SEG11	RB4	0	0	DIG	LATB<4> data output.		
		1	Ι	TTL	PORTB<4> data input; weak pull-up when $\overline{\text{RBPU}}$ bit is cleared.		
	KBI0	1	I	TTL	Interrupt-on-pin change.		
	SEG11	x	0	ANA	LCD segment 11 output; disables all other pin functions.		
RB5/KBI1/SEG29	RB5	0	0	DIG	LATB<5> data output.		
		1	Ι	TTL	PORTB<5> data input; weak pull-up when RBPU bit is cleared.		
	KBI1	1	Ι	TTL	Interrupt-on-pin change.		
	SEG29	x	0	ANA	LCD segment 29 output; disables all other pin functions.		
RB6/KBI2/PGC	RB6	0	0	DIG	LATB<6> data output.		
		1	Ι	TTL	PORTB<6> data input; weak pull-up when RBPU bit is cleared.		
	KBI2	1	I	TTL	Interrupt-on-pin change.		
	PGC	x	Ι	ST	Serial execution (ICSP™) clock input for ICSP and ICD operation.		
RB7/KBI3/PGD	RB7	0	0	DIG	LATB<7> data output.		
		1	Ι	TTL	PORTB<7> data input; weak pull-up when RBPU bit is cleared.		
	KBI3	1	Ι	TTL	Interrupt-on-pin change.		
	PGD	x	0	DIG	Serial execution data output for ICSP and ICD operation.		
		x	I	ST	Serial execution data input for ICSP and ICD operation.		

TABLE 9-5: PORTB FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input,

 $\mathsf{TTL} = \mathsf{TTL} \text{ Buffer Input, } \mathbf{x} = \mathsf{Don't} \text{ care (TRIS bit does not affect port direction or is overridden for this option)}.$

TABLE 9-6:	SUMMARY OF REGISTERS ASSOCIATED WITH PORTB
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Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	55
LATB	LATB7	LATB6	LATB5	LATB4	LATB3	LATB2	LATB1	LATB0	54
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	54
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51
INTCON2	RBPU	INTEDG0	INTEDG1	INTEDG2	INTEDG3	TMR0IP	INT3IP	RBIP	51
INTCON3	INT2IP	INT1IP	INT3IE	INT2IE	INT1IE	INT3IF	INT2IF	INT1IF	51
LCDSE1	SE15	SE14	SE13	SE12	SE11	SE10	SE09	SE08	53
LCDSE3	SE31	SE30	SE29	SE28	SE27	SE26	SE25	SE24	53

Legend: Shaded cells are not used by PORTB.

9.4 PORTC, TRISC and LATC Registers

PORTC is an 8-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISC and LATC. Only PORTC pins, RC2 through RC7, are digital only pins and can tolerate input voltages up to 5.5V.

PORTC is multiplexed with CCP, MSSP and EUSART peripheral functions (Table 9-7). The pins have Schmitt Trigger input buffers. The pins for CCP, SPI and EUSART are also configurable for open-drain output whenever these functions are active. Open-drain configuration is selected by setting the SPIOD, CCPxOD, and U1OD control bits (TRISG<7:5> and LATG<6>, respectively).

RC1 is normally configured as the default peripheral pin for the CCP2 module. Assignment of CCP2 is controlled by Configuration bit, CCP2MX (default state, CCP2MX = 1).

When enabling peripheral functions, care should be taken in defining TRIS bits for each PORTC pin. Some peripherals override the TRIS bit to make a pin an output, while other peripherals override the TRIS bit to make a pin an input. The user should refer to the corresponding peripheral section for the correct TRIS bit settings.

Note: These pins are configured as digital inputs on any device Reset. The contents of the TRISC register are affected by peripheral overrides. Reading TRISC always returns the current contents, even though a peripheral device may be overriding one or more of the pins.

RC<7:1> pins are multiplexed with LCD segment drives, controlled by bits in the LCDSE1, LCDSE2, LCDSE3 and LCDSE4 registers. I/O port functionality is only available when the LCD segments are disabled.

	LE 9-3:	INITIALIZING PORTC
CLRF	PORTC	; Initialize PORTC by ; clearing output : data latches
CLRF	LATC	; Alternate method ; to clear output ; data latches
MOVLW	0CFh	; Value used to ; initialize data : direction
MOVWF	TRISC	,

Pin Name	Function	TRIS Setting	I/O	I/O Type	Description		
RC0/T1OSO/	RC0	0	0	DIG	LATC<0> data output.		
T13CKI		1	PORTC<0> data input.				
	T1OSO	x	0	ANA	Timer1 oscillator output; enabled when Timer1 oscillator enabled. Disables dig ital I/O and LCD segment driver.		
	T13CKI	1	Ι	ST	Timer1/Timer3 counter input.		
RC1/T1OSI/							
CCP2/SEG32		1		ST	PORTC<1> data input.		
	T10SI	x		ANA	Timer1 oscillator input.		
	CCP2 ⁽¹⁾	0	0	DIG	CCP2 Compare/PWM output.		
		1		ST	CCP2 Capture input.		
	SEG32	x	0	ANA	LCD segment 32 output; disables all other pin functions.		
RC2/CCP1/	RC2	0	0	DIG	LATC<2> data output.		
SEG13		1		ST	PORTC<2> data input.		
	CCP1	0	0	DIG	CCP1 Compare/PWM output; takes priority over port data.		
		1	Ι	ST	CCP1 Capture input.		
	SEG13	x	0	ANA	LCD segment 13 output; disables all other pin functions.		
RC3/SCK/SCL/	RC3	0	0	DIG	LATC<3> data output.		
SEG17		1	Ι	ST	PORTC<3> data input.		
	SCK	0	0	DIG	SPI clock output (MSSP module); takes priority over port data.		
		1	Ι	ST	SPI clock input (MSSP module).		
	SCL	0	0	DIG	I ² C™ clock output (MSSP module); takes priority over port data.		
		1	Ι	I2C	I ² C clock input (MSSP module); input type depends on module setting.		
	SEG17	x	0	ANA	LCD segment 17 output; disables all other pin functions.		
RC4/SDI/SDA/	RC4	0	0	DIG	LATC<4> data output.		
SEG16		1	Ι	ST	PORTC<4> data input.		
	SDI		Ι	ST	SPI data input (MSSP module).		
	SDA	1	0	DIG	I ² C data output (MSSP module); takes priority over port data.		
		1	Ι	I2C	I ² C data input (MSSP module); input type depends on module setting.		
	SEG16	x	0	ANA	LCD segment 16 output; disables all other pin functions.		
RC5/SDO/	RC5	0	0	DIG	LATC<5> data output.		
SEG12		1	I	ST	PORTC<5> data input.		
	SDO	0	0	DIG	SPI data output (MSSP module).		
	SEG12	x	0	ANA	LCD segment 12 output; disables all other pin functions.		
RC6/TX1/CK1/	RC6	0	0	DIG	LATC<6> data output.		
SEG27		1	I	ST	PORTC<6> data input.		
	TX1	1	0	DIG	Synchronous serial data output (EUSART module); takes priority over port data.		
	CK1	1	0	DIG	Synchronous serial data input (EUSART module); user must configure as an inpu		
		1	I	ST	Synchronous serial clock input (EUSART module).		
	SEG27	x	0	ANA	LCD segment 27 output; disables all other pin functions.		
RC7/RX1/DT1/	RC7	0	0	DIG	LATC<7> data output.		
SEG28		1	Ι	ST	PORTC<7> data input.		
	RX1	1	I	ST	Asynchronous serial receive data input (EUSART module).		
	DT1	1	0	DIG	Synchronous serial data output (EUSART module); takes priority over port data.		
		1		ST	Synchronous serial data input (EUSART module); user must configure as an inpu		

TABLE 9-7:PORTC FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, TTL = TTL Buffer Input, I2C = I²C/SMBus Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Note 1: Default assignment for CCP2 when CCP2MX Configuration bit is set.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page	
PORTC	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0	55	
LATC	LATC7	LATBC6	LATC5	LATCB4	LATC3	LATC2	LATC1	LATC0	54	
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	54	
LATG	U2OD	U10D	_	LATG4	LATG3	LATG2	LATG1	LATG0	54	
TRISG	SPIOD	CCP2OD	CCP10D	TRISG4	TRISG3	TRISG2	TRISG1	TRISG0	54	
LCDSE1	SE15	SE14	SE13	SE12	SE11	SE10	SE09	SE08	53	
LCDSE2	SE23	SE22	SE21	SE20	SE19	SE18	SE17	SE16	53	
LCDSE3	SE31	SE30	SE29	SE28	SE27	SE26	SE25	SE24	53	
LCDSE4	SE39 ⁽¹⁾	SE38 ⁽¹⁾	SE37 ⁽¹⁾	SE36 ⁽¹⁾	SE35 ⁽¹⁾	SE34 ⁽¹⁾	SE33 ⁽¹⁾	SE32	53	

 TABLE 9-8:
 SUMMARY OF REGISTERS ASSOCIATED WITH PORTC

Legend: Shaded cells are not used by PORTC.

Note 1: Unimplemented on 64-pin devices, read as '0'.

9.5 PORTD, TRISD and LATD Registers

PORTD is an 8-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISD and LATD. All pins on PORTD are digital only and tolerate voltages up to 5.5V.

All pins on PORTD are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output.

Note:	These pins are configured as digital inputs
	on any device Reset.

Each of the PORTD pins has a weak internal pull-up. A single control bit can turn off all the pull-ups. This is performed by setting bit RDPU (PORTG<7>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on all device Resets.

All of the PORTD pins are multiplexed with LCD segment drives, controlled by bits in the LCDSE0 register. I/O port functionality is only available when the LCD segments are disabled.

EXAMPLE 9-4: INITIALIZING PORTD

	•	
CLRF	PORTD	; Initialize PORTD by ; clearing output
CLRF	LATD	; data latches ; Alternate method
		; to clear output ; data latches
MOVLW	0CFh	; Value used to ; initialize data : direction
MOVWF	TRISD	; GIFECION ; Set RD<3:0> as inputs ; RD<5:4> as outputs ; RD<7:6> as inputs

IADLE 9-9:	PURIDI		113				
Pin Name	Function	TRIS Setting	I/O	l/O Type	Description		
RD0/SEG0	RD0	0	0	DIG	LATD<0> data output.		
		1	I	ST	PORTD<0> data input.		
	SEG0	x	0	ANA	LCD segment 0 output; disables all other pin functions.		
RD1/SEG1	RD1	0	0	DIG	LATD<1> data output.		
		1	I	ST	PORTD<1> data input.		
	SEG1	x	0	ANA	LCD segment 1 output; disables all other pin functions.		
RD2/SEG2			LATD<2> data output.				
		1	I	ST	PORTD<2> data input.		
	SEG2 x O ANA LCD segment 2 output; disables a		LCD segment 2 output; disables all other pin functions.				
RD3/SEG3	RD3	0	0	DIG	LATD<3> data output.		
		1	I	ST	PORTD<3> data input.		
	SEG3	x	0	ANA	LCD segment 3 output; disables all other pin functions.		
RD4/SEG4	RD4	0	0	DIG	LATD<4> data output.		
		1	I	ST	PORTD<4> data input.		
	SEG4	x	0	ANA	LCD segment 4 output; disables all other pin functions.		
RD5/SEG5	RD5	0	0	DIG	LATD<5> data output.		
		1	I	ST	PORTD<5> data input.		
	SEG5	x	0	ANA	LCD segment 5 output; disables all other pin functions.		
RD6/SEG6	RD6	0	0	DIG	LATD<6> data output.		
		1	I	ST	PORTD<6> data input.		
	SEG6	x	0	ANA	LCD segment 6 output; disables all other pin functions.		
RD7/SEG7	RD7	0	0	DIG	LATD<7> data output.		
		1	Ι	ST	PORTD<7> data input.		
	SEG7	x	I	ANA	LCD segment 7 output; disables all other pin functions.		

TABLE 9-9: PORTD FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

TABLE 9-10:	SUMMARY OF REGISTERS ASSOCIATED WITH PORTD
-------------	--

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
PORTD	RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0	55
LATD	LATD7	LATD6	LATD5	LATD4	LATD3	LATD2	LATD1	LATD0	54
TRISD	TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0	54
PORTG	RDPU	REPU	RJPU ⁽¹⁾	RG4	RG3	RG2	RG1	RG0	54
LCDSE0	SE7	SE6	SE5	SE4	SE3	SE2	SE1	SE0	53

Legend: Shaded cells are not used by PORTD.

Note 1: Unimplemented on 64-pin devices, read as '0'.

9.6 PORTE, TRISE and LATE Registers

PORTE is a 7-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISE and LATE. All pins on PORTE are digital only and tolerate voltages up to 5.5V.

All pins on PORTE are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output. The RE7 pin is also configurable for open-drain output when CCP2 is active on this pin. Open-drain configuration is selected by setting the CCP2OD control bit (TRISG<6>)

Note:	These pins are configured as digital inputs
	on any device Reset.

Each of the PORTE pins has a weak internal pull-up. A single control bit can turn off all the pull-ups. This is performed by setting bit, REPU (PORTG<6>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on any device Reset.

Pins RE6:RE3 are multiplexed with the LCD common drives. I/O port functions are only available on those PORTE pins depending on which commons are active. The configuration is determined by the LMUX1:LMUX0 control bits (LCDCON<1:0>). The availability is summarized in Table 9-11.

TABLE 9-11: PORTE PINS AVAILABLE IN DIFFERENT LCD DRIVE CONFIGURATIONS

LCDCON <1:0>	Active LCD Commons	PORTE Available for I/O
00	COM0	RE6, RE5, RE4
01	COM0, COM1	RE6, RE5
10	COM0, COM1 and COM2	RE6
11	All (COM0 through COM3)	None

Pins RE1 and RE0 are multiplexed with the functions of LCDBIAS2 and LCDBIAS1. When LCD bias generation is required (i.e., any application where the device is connected to an external LCD), these pins cannot be used as digital I/O.

Note:	The pin corresponding to RE2 of other									
	PIC18F parts has the function of									
	LCDBIAS3 in this device. It cannot be used									
	as digital I/O.									

RE7 is multiplexed with LCD segment drive (SEG31) controlled by the LCDSE3<7> bit. I/O port function is only available when the segment is disabled.

RE7 can also be configured as the alternate peripheral pin for the CCP2 module. This is done by clearing the CCP2MX Configuration bit.

EXAMPLE 9-5: INITIALIZING PORTE

CLRF	PORTE	; Initialize PORTE by ; clearing output
CLRF	LATE	; data latches ; Alternate method ; to clear output
MOVLW	03h	; data latches ; Value used to
MOVWF	TRISE	; initialize data ; direction ; Set RE<1:0> as inputs ; RE<7:2> as outputs
		-

ADLE 9-12.	FORTEFUNCTIONS										
Pin Name	Function	TRIS Setting	I/O	l/O Type	Description						
RE0/LCDBIAS1	RE0	0	0	DIG	LATE<0> data output.						
		1	I	ST	PORTE<0> data input.						
	LCDBIAS1	_	Ι	ANA	LCD module bias voltage input.						
RE1/LCDBIAS2	S2 RE1 0 O DIG LATE<1> data output. 1 I ST PORTE<1> data input.				LATE<1> data output.						
		1	Ι	ST	PORTE<1> data input.						
	LCDBIAS2	—	Ι	ANA	LCD module bias voltage input.						
RE3/COM0 RE3 0 O DIG LA		DIG	LATE<3> data output.								
		1	-	ST	PORTE<3> data input.						
	COM0	x	0	ANA	LCD Common 0 output; disables all other outputs.						
RE4/COM1	RE4	0	0	DIG	LATE<4> data output.						
		1	-	ST	PORTE<4> data input.						
	COM1	x	0	ANA	LCD Common 1 output; disables all other outputs.						
RE5/COM2	RE5	0	0	DIG	LATE<5> data output.						
		1	-	ST	PORTE<5> data input.						
	COM2	x	0	ANA	LCD Common 2 output; disables all other outputs.						
RE6/COM3	RE6	0	0	DIG	LATE<6> data output.						
		1	Ι	ST	PORTE<6> data input.						
	COM3	x	0	ANA	LCD Common 3 output; disables all other outputs.						
RE7/CCP2/	RE7	0	0	DIG	LATE<7> data output.						
SEG31		1	Ι	ST	PORTE<7> data input.						
	CCP2 ⁽¹⁾	0	0	DIG	CCP2 Compare/PWM output; takes priority over port data.						
		1	Ι	ST	CCP2 Capture input.						
	SEG31	x	0	ANA	Segment 31 analog output for LCD; disables digital output.						

TABLE 9-12:PORTE FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Note 1: Alternate assignment for CCP2 when CCP2MX Configuration bit is cleared.

TABLE 9-13: SUMMARY OF REGISTERS ASSOCIATED WITH PORTE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
PORTE	RE7	RE6	RE5	RE4	RE3		RE1	RE0	55
LATE	LATE7	LATE6	LATE5	LATE4	LATE3	_	LATE1	LATE0	54
TRISE	TRISE7	TRISE6	TRISE5	TRISE4	TRISE3	_	TRISE1	TRISE0	54
PORTG	RDPU	REPU	RJPU ⁽¹⁾	RG4	RG3	RG2	RG1	RG0	54
TRISG	SPIOD	CCP2OD	CCP10D	TRISG4	TRISG3	TRISG2	TRISG1	TRISG0	54
LCDCON	LCDEN	SLPEN	WERR	_	CS1	CS0	LMUX1	LMUX0	53
LCDSE3	SE31	SE30	SE29	SE28	SE27	SE26	SE25	SE24	53

Legend: Shaded cells are not used by PORTE.

Note 1: Unimplemented on 64-pin devices, read as '0'.

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9.7 PORTF, LATF and TRISF Registers

PORTF is a 7-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISF and LATF. All pins on PORTF are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output.

PORTF is multiplexed with analog peripheral functions, as well as LCD segments. Pins RF1 through RF6 may be used as comparator inputs or outputs by setting the appropriate bits in the CMCON register. To use RF6:RF3 as digital inputs, it is also necessary to turn off the comparators.

- Note 1: On device Resets, pins RF6:RF1 are configured as analog inputs and are read as '0'.
 - 2: To configure PORTF as digital I/O, turn off comparators and set ADCON1 value.

PORTF is also multiplexed with LCD segment drives controlled by bits in the LCDSE2 and LCDSE3 registers. I/O port functions are only available when the segments are disabled.

EXAMPLE 9-6: INITIALIZING PORTF

CLRF	PORTF	; Initialize PORTF by ; clearing output
		; data latches
CLRF	LATF	; Alternate method
		; to clear output
		; data latches
MOVLW	07h	;
MOVWF	CMCON	; Turn off comparators
MOVLW	0Fh	;
MOVWF	ADCON1	; Set PORTF as digital I/O
MOVLW	0CEh	; Value used to
		; initialize data
		; direction
MOVWF	TRISF	; Set RF3:RF1 as inputs
		; RF5:RF4 as outputs
		; RF7:RF6 as inputs

PIC18F85J90 FAMILY

TABLE 9-14: PORTF FUNCTIONS									
Pin Name	Function	TRIS Setting	I/O	l/O Type	Description				
RF1/AN6/C2OUT/	RF1	0	0	DIG	LATF<1> data output; not affected by analog input.				
SEG19		1	Ι	ST	PORTF<1> data input; disabled when analog input enabled.				
	AN6	1	Ι	ANA	A/D input channel 6. Default configuration on POR.				
	C2OUT	0	0	DIG	Comparator 2 output; takes priority over port data.				
	SEG19	x	0	ANA	LCD segment 19 output; disables all other pin functions.				
RF2/AN7/C1OUT/	RF2	LATF<2> data output; not affected by analog input.							
SEG20		1	Ι	ST	PORTF<2> data input; disabled when analog input enabled.				
	AN7	1	I	ANA	A/D input channel 7. Default configuration on POR.				
	C1OUT	0	Comparator 1 output; takes priority over port data.						
	SEG20	x	0	ANA	LCD segment 20 output; disables all other pin functions.				
RF3/AN8/SEG21	RF3	0	0	DIG	LATF<3> data output; not affected by analog input.				
		1	I	ST	PORTF<3> data input; disabled when analog input enabled.				
	AN8	1	I	ANA	A/D input channel 8 and Comparator C2+ input. Default input configuration on POR; not affected by analog output.				
	SEG21	x	0	ANA	LCD segment 21 output; disables all other pin functions.				
RF4/AN9/SEG22	RF4	0	0	DIG	LATF<4> data output; not affected by analog input.				
		1	I	ST	PORTF<4> data input; disabled when analog input enabled.				
	AN9	1	I	ANA	A/D input channel 9 and Comparator C2- input. Default input configuration on POR; does not affect digital output.				
	SEG22	x	0	ANA	LCD segment 22 output; disables all other pin functions.				
RF5/AN10/CVREF/ SEG23	RF5	0	0	DIG	LATF<5> data output; not affected by analog input. Disabled whe CVREF output enabled.				
		1	I	ST	PORTF<5> data input; disabled when analog input enabled. Disabled when CVREF output enabled.				
	AN10	1	I	ANA	A/D input channel 10 and Comparator C1+ input. Default input configuration on POR.				
	CVREF	х	0	ANA	Comparator voltage reference output. Enabling this feature disables digital I/O.				
	SEG23	x	0	ANA	LCD segment 23 output; disables all other pin functions.				
RF6/AN11/SEG24	RF6	0	0	DIG	LATF<6> data output; not affected by analog input.				
		1	I	ST	PORTF<6> data input; disabled when analog input enabled.				
	AN11	1	I	ANA	A/D input channel 11 and Comparator C1- input. Default input configuration on POR; does not affect digital output.				
	SEG24	х	0	ANA	LCD segment 24 output; disables all other pin functions.				
RF7/AN5/SS/	RF7	0	0	DIG	LATF<7> data output; not affected by analog input.				
SEG25		1	Ι	ST	PORTF<7> data input; disabled when analog input enabled.				
	AN5	1	I	ANA	A/D input channel 5. Default configuration on POR.				
	SS	1	I	TTL	Slave select input for MSSP module.				
	SEG25	x	0	ANA	LCD segment 25 output; disables all other pin functions.				
		-			0 • • • • • • • • • • • • • • • • • • •				

TABLE 9-14:PORTF FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, TTL = TTL Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
PORTF	RF7	RF6	RF5	RF4	RF3	RF2	RF1		54
LATF	LATF7	LATF6	LATF5	LATF4	LATF3	LATF2	LATF1	—	54
TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1	—	54
ADCON1	—	_	VCFG1	VCFG0	PCFG3	PCFG2	PCFG1	PCFG0	53
CMCON	C2OUT	C10UT	C2INV	C1INV	CIS	CM2	CM1	CM0	53
CVRCON	CVREN	CVROE	CVRR	CVRSS	CVR3	CVR2	CVR1	CVR0	53
LCDSE2	SE23	SE22	SE21	SE20	SE19	SE18	SE17	SE16	53
LCDSE3	SE31	SE30	SE29	SE28	SE27	SE26	SE25	SE24	53

TABLE 9-15: SUMMARY OF REGISTERS ASSOCIATED WITH PORTF

Legend: — = unimplemented, read as '0'. Shaded cells are not used by PORTF.

9.8 PORTG, TRISG and LATG Registers

PORTG is a 5-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISG and LATG. All pins on PORTG are digital only and tolerate voltages up to 5.5V.

PORTG is multiplexed with both AUSART and LCD functions (Table 9-16). When operating as I/O, all PORTG pins have Schmitt Trigger input buffers. The RG1 pin is also configurable for open-drain output when the AUSART is active. Open-drain configuration is selected by setting the U2OD control bit (LATG<7>).

RG4 is multiplexed with LCD segment drives controlled by bits in the LCDSE2 register. The I/O port function is only available when the segments are disabled.

RG3 and RG2 are multiplexed with VLCAP pins for the LCD charge pump, and RG0 is multiplexed with LCDBIAS0 bias voltage input. When these pins are used for LCD bias generation, the I/O and other functions are unavailable.

When enabling peripheral functions, care should be taken in defining TRIS bits for each PORTG pin. Some peripherals override the TRIS bit to make a pin an output, while other peripherals override the TRIS bit to make a pin an input. The user should refer to the corresponding peripheral section for the correct TRIS bit settings. The pin override value is not loaded into the TRIS register. This allows read-modify-write of the TRIS register without concern due to peripheral overrides. Although the port itself is only five bits wide, the PORTG<7:5> bits are still implemented to control the weak pull-ups on the I/O ports associated with PORTD, PORTE and PORTJ. Setting these bits enables the respective port pull-ups.

Most of the corresponding TRISG and LATG bits are implemented as open-drain control bits for CCP1, CCP2 and SPI (TRISG<7:5>), and the USARTs (LATG<7:6>). Setting these bits configures the output pin for the corresponding peripheral for open-drain operation. LATG<5> is not implemented.

EXAMPLE 9-7: INITIALIZING PORTG

CLRF	PORTG	; Initialize PORTG by ; clearing output
CLRF	LATG	; data latches ; Alternate method ; to clear output
MOVLW	04h	; data latches ; Value used to ; initialize data
MOVWF	TRISG	; direction ; Set RG1:RG0 as outputs ; RG2 as input ; RG4:RG3 as inputs

TADEE 5-10.									
Pin Name	Function	TRIS Setting	I/O	I/O Type	Description				
RG0/LCDBIAS0	RG0	0	0	DIG	LATG<0> data output.				
		1	I	ST	PORTG<0> data input.				
	LCDBIAS0	x	I	ANA	LCD module bias voltage input.				
RG1/TX2/CK2	RG1	0	0	DIG	LATG<1> data output.				
		1	Ι	ST	PORTG<1> data input.				
	TX2	1	0	DIG	Synchronous serial data output (AUSART module); takes priority over port data.				
	CK2	1	0	DIG	Synchronous serial data input (AUSART module); user must configure as an input.				
		1	I	ST	Synchronous serial clock input (AUSART module).				
RG2/RX2/DT2/V	RG2	0	0	DIG	LATG<2> data output.				
LCAP1		1	I	ST	PORTG<2> data input.				
	RX2	1	Ι	ST	Asynchronous serial receive data input (AUSART module).				
	DT2	1	0	DIG	Synchronous serial data output (AUSART module); takes priority over port data.				
		1	Ι	ST	Synchronous serial data input (AUSART module); user must configure as an input.				
	VLCAP1	x	I	ANA	LCD charge pump capacitor input.				
RG3/VLCAP2	RG3	0	0	DIG	LATG<3> data output.				
		1	Ι	ST	PORTG<3> data input.				
	VLCAP2	x	I	ANA	LCD charge pump capacitor input.				
RG4/SEG26	RG4	0	0	DIG	LATG<4> data output.				
		1	I	ST	PORTG<4> data input.				
	SEG26	x	0	ANA	LCD segment 26 output; disables all other pin functions.				

TABLE 9-16:PORTG FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

TABLE 9-17: SUMMARY OF REGISTERS ASSOCIATED WITH PORTG

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
PORTG	RDPU	REPU	RJPU ⁽¹⁾	RG4	RG3	RG2	RG1	RG0	54
LATG	U2OD	U10D	—	LATG4	LATG3	LATG2	LATG1	LATG0	54
TRISG	SPIOD	CCP2OD	CCP10D	TRISG4	TRISG3	TRISG2	TRISG1	TRISG0	54
LCDSE3	SE31	SE30	SE29	SE28	SE27	SE26	SE25	SE24	53

Legend: — = unimplemented, read as '0'. Shaded cells are not used by PORTG.

Note 1: Unimplemented on 64-pin devices, read as '0'.

9.9 PORTH, LATH and TRISH Registers

Note:	PORTH	is	available	only	on	80-pin
	devices.					

PORTH is an 8-bit wide, bidirectional I/O port. The corresponding Data Direction and Output Latch registers are TRISH and LATH. All pins are digital only and tolerate voltages up to 5.5V.

All pins on PORTH are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output.

All PORTH pins are multiplexed with LCD segment drives controlled by the LCDSE5 register. I/O port functions are only available when the segments are disabled.

EXAMP	LE 9-8:	INITIALIZING PORTH
CLRF	PORTH	; Initialize PORTH by
		; clearing output
		; data latches
CLRF	LATH	; Alternate method
		; to clear output
		; data latches
MOVLW	0Fh	; Configure PORTH as
MOVWF	ADCON1	; digital I/O
MOVLW	0CFh	; Value used to
		; initialize data
		; direction
MOVWF	TRISH	; Set RH3:RH0 as inputs
		; RH5:RH4 as outputs
		; RH7:RH6 as inputs

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Pin Name	Function	TRIS Setting	I/O	I/O Type	Description	
RH0/SEG47	RH0	0	O DIG LATH<0> data output.		LATH<0> data output.	
		1	I	ST	PORTH<0> data input.	
	SEG47	x	0	ANA	LCD segment 47 output; disables all other pin functions.	
RH1/SEG46	RH1	0	0	DIG	LATH<1> data output.	
		1	I	ST	PORTH<1> data input.	
	SEG46	x	0	ANA	LCD segment 46 output; disables all other pin functions.	
RH2/SEG45	RH2	0	0	DIG	LATH<2> data output.	
		1	I	ST	PORTH<2> data input.	
	SEG45	x	0	ANA	LCD segment 45 output; disables all other pin functions.	
RH3/SEG44	RH3	0	0	DIG	LATH<3> data output.	
		1	Ι	ST	PORTH<3> data input.	
	SEG44	x	0	ANA	LCD segment 44 output; disables all other pin functions.	
RH4/SEG40	RH4	0	0	DIG	LATH<4> data output.	
		1	Ι	ST	PORTH<4> data input.	
	SEG40	x	0	ANA	LCD segment 40 output; disables all other pin functions.	
RH5/SEG41	RH5	0	0	DIG	LATH<5> data output.	
		1	Ι	ST	PORTH<5> data input.	
	SEG41	x	0	ANA	LCD segment 41 output; disables all other pin functions.	
RH6/SEG42	RH6	0	0	DIG	LATH<6> data output.	
		1	I	ST	PORTH<6> data input.	
	SEG42	x	0	ANA	LCD segment 42 output; disables all other pin functions.	
RH7/SEG43	RH7	0	0	DIG	LATH<7> data output.	
		1	I	ST	PORTH<7> data input.	
	SEG43	x	0	ANA	LCD segment 43 output; disables all other pin functions.	

TABLE 9-18: PORTH FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
PORTH	RH7	RH6	RH5	RH4	RH3	RH2	RH1	RH0	54
LATH	LATH7	LATH6	LATH5	LATH4	LATH3	LATH2	LATH1	LATH0	54
TRISH	TRISH7	TRISH6	TRISH5	TRISH4	TRISH3	TRISH2	TRISH1	TRISH0	54
LCDSE5	SE47	SE46	SE45	SE44	SE43	SE42	SE41	SE40	53

9.10 PORTJ, TRISJ and LATJ Registers

Note: PORTJ is available only on 80-pin devices.

PORTJ is an 8-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISJ and LATJ. All pins on PORTJ are digital only and tolerate voltages up to 5.5V.

All pins on PORTJ are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output.

Note: These pins are configured as digital inputs on any device Reset.

All PORTJ pins except RJ0 are multiplexed with LCD segment drives controlled by the LCDSE4 register. I/O port functions are only available on these pins when the segments are disabled.

Each of the PORTJ pins has a weak internal pull-up. The pull-ups are provided to keep the inputs at a known state for the external memory interface while powering up. A single control bit can turn off all the pull-ups. This is performed by clearing bit RJPU (PORTG<5>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on any device Reset.

EXAMPLE 9-9:	INITIALIZING PORTJ

CLRF	PORTJ	; Initialize PORTJ by
		; clearing output latches
CLRF	LATJ	; Alternate method
		; to clear output latches
MOVLW	0CFh	; Value used to
		; initialize data
		; direction
MOVWF	TRISJ	; Set RJ3:RJ0 as inputs
		; RJ5:RJ4 as output
		; RJ7:RJ6 as inputs

Pin Name	Function	TRIS Setting	I/O	l/O Type	Description
RJO RJO		0	0	DIG	LATJ<0> data output.
		1	I	ST	PORTJ<0> data input.
RJ1/SEG33	RJ1	0	0	DIG	LATJ<1> data output.
		1	Ι	ST	PORTJ<1> data input.
	SEG33	x	0	ANA	LCD segment 33 output; disables all other pin functions.
RJ2/SEG34	RJ2	0	0	DIG	LATJ<2> data output.
		1	-	ST	PORTJ<2> data input.
	SEG34	x	0	ANA	LCD segment 34 output; disables all other pin functions.
RJ3/SEG35	RJ3	0	0	DIG	LATJ<3> data output.
		1	-	ST	PORTJ<3> data input.
	SEG35	x	0	ANA	LCD segment 35 output; disables all other pin functions.
RJ4/SEG39	RJ4	0	0	DIG	LATJ<4> data output.
		1	-	ST	PORTJ<4> data input.
	SEG39	x	0	ANA	LCD segment 39 output; disables all other pin functions.
RJ5/SEG38	RJ5	0	0	DIG	LATJ<5> data output.
		1	-	ST	PORTJ<5> data input.
	SEG38	x	0	ANA	LCD segment 38 output; disables all other pin functions.
RJ6/SEG37	RJ6	0	0	DIG	LATJ<6> data output.
		1	-	ST	PORTJ<6> data input.
	SEG37	x	0	ANA	LCD segment 37 output; disables all other pin functions.
RJ7/SEG36	RJ7	0	0	DIG	LATJ<7> data output.
		1	Ι	ST	PORTJ<7> data input.
	SEG36	x	0	ANA	LCD segment 36 output; disables all other pin functions.

TABLE 9-20: PORTJ FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

TABLE 9-21. SUMIMANT OF REGISTERS ASSOCIATED WITH FURT	TABLE 9-21:	SUMMARY OF REGISTERS ASSOCIATED WITH PORTJ
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Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
PORTJ	RJ7	RJ6	RJ5	RJ4	RJ3	RJ2	RJ1	RJ0	54
LATJ	LATJ7	LATJ6	LATJ5	LATJ4	LATJ3	LATJ2	LATJ1	LATJ0	54
TRISJ	TRISJ7	TRISJ6	TRISJ5	TRISJ4	TRISJ3	TRISJ2	TRISJ1	TRISJ0	54
PORTG	RDPU	REPU	RJPU ⁽¹⁾	RG4	RG3	RG2	RG1	RG0	54
LCDSE4	SE39	SE38	SE37	SE36	SE35	SE34	SE33	SE32	53

Legend: Shaded cells are not used by PORTJ.

Note 1: Unimplemented on 64-pin devices, read as '0'.

10.0 TIMER0 MODULE

The Timer0 module incorporates the following features:

- Software selectable operation as a timer or counter in both 8-bit or 16-bit modes
- Readable and writable registers
- Dedicated 8-bit, software programmable
 prescaler
- · Selectable clock source (internal or external)
- Edge select for external clock
- Interrupt-on-overflow

The T0CON register (Register 10-1) controls all aspects of the module's operation, including the prescale selection; it is both readable and writable.

A simplified block diagram of the Timer0 module in 8-bit mode is shown in Figure 10-1. Figure 10-2 shows a simplified block diagram of the Timer0 module in 16-bit mode.

REGISTER 10-1: TOCON: TIMER0 CONTROL REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
TMR00N	T08BIT	TOCS	TOSE	PSA	T0PS2	T0PS1	T0PS0
bit 7							bit 0

Legend: R = Readal	ble bit	W = Writable bit	U = Unimplemented bit	. read as '0'						
-n = Value		'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown						
bit 7	TMR0ON:	Timer0 On/Off Control bit								
	1 = Enable									
	0 = Stops	Timer0								
bit 6	T08BIT : Ti	imer0 8-Bit/16-Bit Control b	it							
		0 is configured as an 8-bit ti								
	0 = Timer(0 is configured as a 16-bit ti	imer/counter							
bit 5	TOCS: Tim	TOCS: Timer0 Clock Source Select bit								
	1 = Transi	1 = Transition on TOCKI pin								
	0 = Interna	al instruction cycle clock (C	LKO)							
bit 4	T0SE: Timer0 Source Edge Select bit									
	1 = Increm	nent on high-to-low transitio	on on TOCKI pin							
	0 = Increm	nent on low-to-high transitio	on on TOCKI pin							
bit 3	PSA: Time	er0 Prescaler Assignment b	pit							
			. Timer0 clock input bypasses							
	0 = Timer(0 prescaler is assigned. Tim	ner0 clock input comes from pr	escaler output.						
bit 2-0	T0PS2:T0PS0: Timer0 Prescaler Select bits									
		56 Prescale value								
		28 Prescale value								
		4 Prescale value								
		2 Prescale value 6 Prescale value								
		Prescale value								
	001 = 1:4									
	000 = 1:2	Prescale value								

10.1 Timer0 Operation

Timer0 can operate as either a timer or a counter. The mode is selected with the TOCS bit (TOCON<5>). In Timer mode (TOCS = 0), the module increments on every clock by default unless a different prescaler value is selected (see **Section 10.3 "Prescaler**"). If the TMR0 register is written to, the increment is inhibited for the following two instruction cycles. The user can work around this by writing an adjusted value to the TMR0 register.

The Counter mode is selected by setting the T0CS bit (= 1). In this mode, Timer0 increments either on every rising or falling edge of pin RA4/T0CKI. The incrementing edge is determined by the Timer0 Source Edge Select bit, T0SE (T0CON<4>); clearing this bit selects the rising edge. Restrictions on the external clock input are discussed below.

An external clock source can be used to drive Timer0, however, it must meet certain requirements to ensure that the external clock can be synchronized with the internal phase clock (Tosc). There is a delay between synchronization and the onset of incrementing the timer/counter.

10.2 Timer0 Reads and Writes in 16-Bit Mode

TMR0H is not the actual high byte of Timer0 in 16-bit mode. It is actually a buffered version of the real high byte of Timer0, which is not directly readable nor writable (refer to Figure 10-2). TMR0H is updated with the contents of the high byte of Timer0 during a read of TMR0L. This provides the ability to read all 16 bits of Timer0 without having to verify that the read of the high and low byte were valid, due to a rollover between successive reads of the high and low byte.

Similarly, a write to the high byte of Timer0 must also take place through the TMR0H Buffer register. The high byte is updated with the contents of TMR0H when a write occurs to TMR0L. This allows all 16 bits of Timer0 to be updated at once.

FIGURE 10-1: TIMER0 BLOCK DIAGRAM (8-BIT MODE)

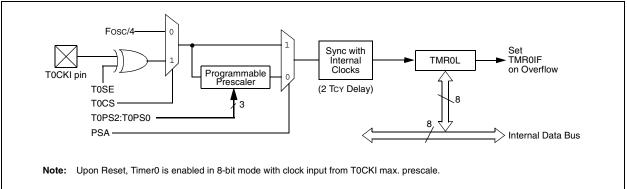
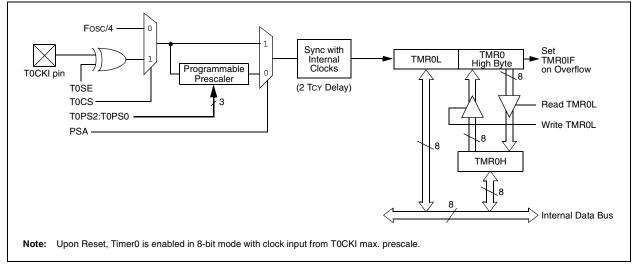


FIGURE 10-2: TIMER0 BLOCK DIAGRAM (16-BIT MODE)



10.3 Prescaler

An 8-bit counter is available as a prescaler for the Timer0 module. The prescaler is not directly readable or writable. Its value is set by the PSA and T0PS2:T0PS0 bits (T0CON<3:0>) which determine the prescaler assignment and prescale ratio.

Clearing the PSA bit assigns the prescaler to the Timer0 module. When it is assigned, prescale values from 1:2 through 1:256, in power-of-2 increments, are selectable.

When assigned to the Timer0 module, all instructions writing to the TMR0 register (e.g., CLRF TMR0, MOVWF TMR0, BSF TMR0, etc.) clear the prescaler count.

Note:	Writing to TMR0 when the prescaler is
	assigned to Timer0 will clear the prescaler
	count but will not change the prescaler
	assignment.

10.3.1 SWITCHING PRESCALER ASSIGNMENT

The prescaler assignment is fully under software control and can be changed "on-the-fly" during program execution.

10.4 Timer0 Interrupt

The TMR0 interrupt is generated when the TMR0 register overflows from FFh to 00h in 8-bit mode, or from FFFFh to 0000h in 16-bit mode. This overflow sets the TMR0IF flag bit. The interrupt can be masked by clearing the TMR0IE bit (INTCON<5>). Before re-enabling the interrupt, the TMR0IF bit must be cleared in software by the Interrupt Service Routine.

Since Timer0 is shut down in Sleep mode, the TMR0 interrupt cannot awaken the processor from Sleep.

TABLE 10-1. MEDISTENS ASSOCIATED WITH INVEND											
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page		
TMR0L	Timer0 Reg	Timer0 Register Low Byte									
TMR0H	Timer0 Reg	Timer0 Register High Byte									
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51		
TOCON	TMR0ON	T08BIT	TOCS	T0SE	PSA	T0PS2	T0PS1	T0PS0	52		
TRISA	TRISA7 ⁽¹⁾	TRISA6 ⁽¹⁾	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	54		

TABLE 10-1: REGISTERS ASSOCIATED WITH TIMER0

Legend: — = unimplemented, read as '0'. Shaded cells are not used by Timer0.

Note 1: RA6/RA7 and their associated latch and direction bits are configured as port pins only when the internal oscillator is selected as the default clock source (FOSC2 Configuration bit = 0); otherwise, they are disabled and these bits read as '0'.

PIC18F85J90 FAMILY

NOTES:

11.0 TIMER1 MODULE

The Timer1 timer/counter module incorporates these features:

- Software selectable operation as a 16-bit timer or counter
- Readable and writable 8-bit registers (TMR1H and TMR1L)
- Selectable clock source (internal or external) with device clock or Timer1 oscillator internal options
- Interrupt on overflow
- Reset on CCP Special Event Trigger
- Device clock status flag (T1RUN)

A simplified block diagram of the Timer1 module is shown in Figure 11-1. A block diagram of the module's operation in Read/Write mode is shown in Figure 11-2.

The module incorporates its own low-power oscillator to provide an additional clocking option. The Timer1 oscillator can also be used as a low-power clock source for the microcontroller in power-managed operation.

Timer1 can also be used to provide Real-Time Clock (RTC) functionality to applications with only a minimal addition of external components and code overhead.

Timer1 is controlled through the T1CON Control register (Register 11-1). It also contains the Timer1 Oscillator Enable bit (T1OSCEN). Timer1 can be enabled or disabled by setting or clearing control bit, TMR1ON (T1CON<0>).

R/W-0	R-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
RD16	T1RUN	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N
bit 7							bit 0

REGISTER 11-1: T1CON: TIMER1 CONTROL REGISTER

Legend:										
R = Reada	ble bit	W = Writable bit	U = Unimplemented bit,	read as '0'						
-n = Value	at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknowr						
bit 7	BD16 : 16	-Bit Read/Write Mode Enab	le bit							
	1 = Enat	les register read/write of TIr les register read/write of Tin	ner1 in one 16-bit operation							
bit 6	T1RUN:	Timer1 System Clock Status	bit							
		ce clock is derived from Time ce clock is derived from anot								
bit 5-4	T1CKPS	I:T1CKPS0: Timer1 Input C	lock Prescale Select bits							
	10 = 1:4 01 = 1:2	Prescale value Prescale value Prescale value Prescale value								
bit 3	T10SCE	T10SCEN: Timer1 Oscillator Enable bit								
	0 = Timei	1 oscillator is enabled 1 oscillator is shut off ator inverter and feedback r	esistor are turned off to elimina	te power drain.						
bit 2		Timer1 External Clock Inpu								
	<u>When TN</u> 1 = Do no 0 = Syncl	H <u>1CS = 1:</u> ot synchronize external clock nronize external clock input	-							
		I <u>R1CS = 0:</u> ignored. Timer1 uses the in	ternal clock when TMR1CS = 0).						
bit 1		: Timer1 Clock Source Selec								
		rnal clock from pin RC0/T1O nal clock (FOSC/4)	SO/T13CKI (on the rising edge	e)						
bit 0		: Timer1 On bit les Timer1 s Timer1								

11.1 Timer1 Operation

Timer1 can operate in one of these modes:

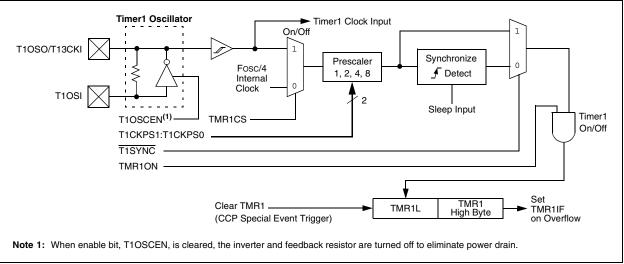
- Timer
- Synchronous Counter
- Asynchronous Counter

The operating mode is determined by the clock select bit, TMR1CS (T1CON<1>). When TMR1CS is cleared (= 0), Timer1 increments on every internal instruction

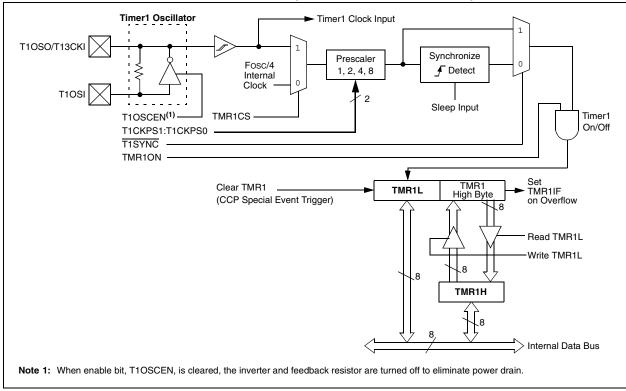
cycle (Fosc/4). When the bit is set, Timer1 increments on every rising edge of the Timer1 external clock input or the Timer1 oscillator, if enabled.

When Timer1 is enabled, the RC1/T1OSI and RC0/T1OSO/T13CKI pins become inputs. This means the values of TRISC<1:0> are ignored and the pins are read as '0'.

FIGURE 11-1: TIMER1 BLOCK DIAGRAM (8-BIT MODE)







11.2 Timer1 16-Bit Read/Write Mode

Timer1 can be configured for 16-bit reads and writes (see Figure 11-2). When the RD16 control bit (T1CON<7>) is set, the address for TMR1H is mapped to a buffer register for the high byte of Timer1. A read from TMR1L will load the contents of the high byte of Timer1 into the Timer1 High Byte Buffer register. This provides the user with the ability to accurately read all 16 bits of Timer1 without having to determine whether a read of the high byte, followed by a read of the low byte, has become invalid due to a rollover between reads.

A write to the high byte of Timer1 must also take place through the TMR1H Buffer register. The Timer1 high byte is updated with the contents of TMR1H when a write occurs to TMR1L. This allows a user to write all 16 bits to both the high and low bytes of Timer1 at once.

The high byte of Timer1 is not directly readable or writable in this mode. All reads and writes must take place through the Timer1 High Byte Buffer register. Writes to TMR1H do not clear the Timer1 prescaler. The prescaler is only cleared on writes to TMR1L.

11.3 Timer1 Oscillator

An on-chip crystal oscillator circuit is incorporated between pins T1OSI (input) and T1OSO (amplifier output). It is enabled by setting the Timer1 Oscillator Enable bit, T1OSCEN (T1CON<3>). The oscillator is a low-power circuit rated for 32 kHz crystals. It will continue to run during all power-managed modes. The circuit for a typical LP oscillator is shown in Figure 11-3. Table 11-1 shows the capacitor selection for the Timer1 oscillator.

The user must provide a software time delay to ensure proper start-up of the Timer1 oscillator.

FIGURE 11-3: EXTERNAL COMPONENTS FOR THE TIMER1 LP OSCILLATOR

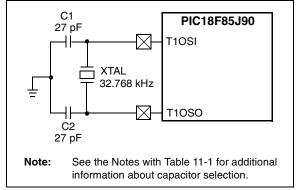


TABLE 11-1: CAPACITOR SELECTION FOR THE TIMER1 OSCILLATOR^(2,3,4)

USUILLATUN								
Oscillator Type	Freq.	C1	C2					
LP	32.768 kHz	27 pF ⁽¹⁾	27 pF ⁽¹⁾					
Note 1: Microchip suggests these values as a starting point in validating the oscillator circuit.								
	Higher capacitance increases the stability of the oscillator but also increases the start-up time.							
	Since each resonator/crystal has its own characteristics, the user should consult the resonator/crystal manufacturer for appropriate values of external components.							
	Capacitor valu only.	es are for des	ign guidance					

11.3.1 USING TIMER1 AS A CLOCK SOURCE

The Timer1 oscillator is also available as a clock source in power-managed modes. By setting the System Clock Select bits, SCS1:SCS0 (OSCCON<1:0>), to '01', the device switches to SEC_RUN mode; both the CPU and peripherals are clocked from the Timer1 oscillator. If the IDLEN bit (OSCCON<7>) is cleared and a SLEEP instruction is executed, the device enters SEC_IDLE mode. Additional details are available in Section 3.0 "Power-Managed Modes".

Whenever the Timer1 oscillator is providing the clock source, the Timer1 system clock status flag, T1RUN (T1CON<6>), is set. This can be used to determine the controller's current clocking mode. It can also indicate the clock source being currently used by the Fail-Safe Clock Monitor. If the Clock Monitor is enabled and the Timer1 oscillator fails while providing the clock, polling the T1RUN bit will indicate whether the clock is being provided by the Timer1 oscillator or another source.

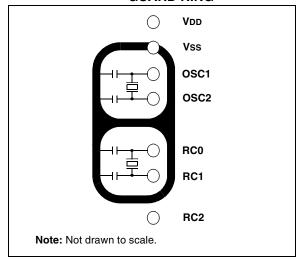
11.3.2 TIMER1 OSCILLATOR LAYOUT CONSIDERATIONS

The Timer1 oscillator circuit draws very little power during operation. Due to the low-power nature of the oscillator, it may also be sensitive to rapidly changing signals in close proximity.

The oscillator circuit, shown in Figure 11-3, should be located as close as possible to the microcontroller. There should be no circuits passing within the oscillator circuit boundaries other than Vss or VDD.

If a high-speed circuit must be located near the oscillator (such as the CCP1 pin in Output Compare or PWM mode, or the primary oscillator using the OSC2 pin), a grounded guard ring around the oscillator circuit, as shown in Figure 11-4, may be helpful when used on a single-sided PCB or in addition to a ground plane.

FIGURE 11-4: OSCILLATOR CIRCUIT WITH GROUNDED GUARD RING



11.4 Timer1 Interrupt

The TMR1 register pair (TMR1H:TMR1L) increments from 0000h to FFFFh and rolls over to 0000h. The Timer1 interrupt, if enabled, is generated on overflow which is latched in interrupt flag bit, TMR1IF (PIR1<0>). This interrupt can be enabled or disabled by setting or clearing the Timer1 Interrupt Enable bit, TMR1IE (PIE1<0>).

11.5 Resetting Timer1 Using the CCP Special Event Trigger

If CCP1 or CCP2 is configured to use Timer1 and to generate a Special Event Trigger in Compare mode (CCPxM3:CCPxM0 = 1011), this signal will reset Timer3. The trigger from CCP2 will also start an A/D conversion if the A/D module is enabled (see **Section 14.3.4 "Special Event Trigger"** for more information).

The module must be configured as either a timer or a synchronous counter to take advantage of this feature. When used this way, the CCPRxH:CCPRxL register pair effectively becomes a period register for Timer1.

If Timer1 is running in Asynchronous Counter mode, this Reset operation may not work.

In the event that a write to Timer1 coincides with a Special Event Trigger, the write operation will take precedence.

Note:	The Special Event Triggers from the CCPx module will not set the TMR1IF interrupt
	flag bit (PIR1<0>).

11.6 Using Timer1 as a Real-Time Clock

Adding an external LP oscillator to Timer1 (such as the one described in **Section 11.3 "Timer1 Oscillator"** above) gives users the option to include RTC functionality to their applications. This is accomplished with an inexpensive watch crystal to provide an accurate time base and several lines of application code to calculate the time. When operating in Sleep mode and using a battery or supercapacitor as a power source, it can completely eliminate the need for a separate RTC device and battery backup.

The application code routine, RTCisr, shown in Example 11-1, demonstrates a simple method to increment a counter at one-second intervals using an Interrupt Service Routine. Incrementing the TMR1 register pair to overflow triggers the interrupt and calls the routine which increments the seconds counter by one. Additional counters for minutes and hours are incremented as the previous counter overflows.

Since the register pair is 16 bits wide, counting up to overflow the register directly from a 32.768 kHz clock would take 2 seconds. To force the overflow at the required one-second intervals, it is necessary to preload it. The simplest method is to set the MSb of TMR1H with a BSF instruction. Note that the TMR1L register is never preloaded or altered; doing so may introduce cumulative error over many cycles.

For this method to be accurate, Timer1 must operate in Asynchronous mode and the Timer1 overflow interrupt must be enabled (PIE1<0> = 1) as shown in the routine, RTCinit. The Timer1 oscillator must also be enabled and running at all times.

EXAMPLE	11-1:	IMPLEMENTIN	G A REAL-TIME CLOCK USING A TIMER1 INTERRUPT SERVICE
RTCinit			
	MOVLW	80h	; Preload TMR1 register pair
	MOVWF	TMR1H	; for 1 second overflow
	CLRF	TMR1L	
	MOVLW	b'00001111'	; Configure for external clock,
	MOVWF	T1CON	; Asynchronous operation, external oscillator
	CLRF	secs	; Initialize timekeeping registers
	CLRF	mins	;
	MOVLW	.12	
	MOVWF	hours	
	BSF	PIE1, TMR1IE	; Enable Timer1 interrupt
	RETURN		
RTCisr			
	BSF	TMR1H, 7	; Preload for 1 sec overflow
	BCF	PIR1, TMR1IF	; Clear interrupt flag
	INCF	secs, F	; Increment seconds
	MOVLW	.59	; 60 seconds elapsed?
	CPFSGT	secs	
	RETURN		; No, done
	CLRF	secs	; Clear seconds
	INCF	mins, F	; Increment minutes
	MOVLW	.59	; 60 minutes elapsed?
	CPFSGT		
	RETURN		; No, done
	CLRF	mins	; clear minutes
	INCF	hours, F	; Increment hours
	MOVLW	.23	; 24 hours elapsed?
	CPFSGT		
	RETURN		; No, done
	CLRF	hours	; Reset hours
	RETURN		; Done

TABLE 11-2: REGISTERS ASSOCIATED WITH TIMER1 AS A TIMER/COUNTER

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51
PIR1	—	ADIF	RC1IF	TX1IF	SSPIF	_	TMR2IF	TMR1IF	54
PIE1	—	ADIE	RC1IE	TX1IE	SSPIE	_	TMR2IE	TMR1IE	54
IPR1	—	ADIP	RC1IP	TX1IP	SSPIP	_	TMR2IP	TMR1IP	54
TMR1L	Timer1 Reg	gister Low By	/te						52
TMR1H	Timer1 Register High Byte								
T1CON	RD16	T1RUN	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N	52

Legend: Shaded cells are not used by the Timer1 module.

PIC18F85J90 FAMILY

NOTES:

12.0 TIMER2 MODULE

The Timer2 module incorporates the following features:

- 8-bit Timer and Period registers (TMR2 and PR2, respectively)
- Readable and writable (both registers)
- Software programmable prescaler (1:1, 1:4 and 1:16)
- Software programmable postscaler (1:1 through 1:16)
- Interrupt on TMR2 to PR2 match
- Optional use as the shift clock for the MSSP module

The module is controlled through the T2CON register (Register 12-1) which enables or disables the timer and configures the prescaler and postscaler. Timer2 can be shut off by clearing control bit, TMR2ON (T2CON<2>), to minimize power consumption.

A simplified block diagram of the module is shown in Figure 12-1.

12.1 Timer2 Operation

In normal operation, TMR2 is incremented from 00h on each clock (Fosc/4). A 4-bit counter/prescaler on the clock input gives direct input, divide-by-4 and divide-by-16 prescale options. These are selected by the prescaler control bits, T2CKPS1:T2CKPS0 (T2CON<1:0>). The value of TMR2 is compared to that of the Period register, PR2, on each clock cycle. When the two values match, the comparator generates a match signal as the timer output. This signal also resets the value of TMR2 to 00h on the next cycle and drives the output counter/postscaler (see Section 12.2 "Timer2 Interrupt").

The TMR2 and PR2 registers are both directly readable and writable. The TMR2 register is cleared on any device Reset, while the PR2 register initializes at FFh. Both the prescaler and postscaler counters are cleared on the following events:

- a write to the TMR2 register
- a write to the T2CON register
- any device Reset (Power-on Reset, MCLR Reset, Watchdog Timer Reset or Brown-out Reset)

TMR2 is not cleared when T2CON is written.

REGISTER 12-1: T2CON: TIMER2 CONTROL REGISTER

U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	T2OUTPS3	T2OUTPS2	T2OUTPS1	T2OUTPS0	TMR2ON	T2CKPS1	T2CKPS0
bit 7							bit 0

Legend:					
R = Readable bit	R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'				
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown		

bit 7 bit 6-3	Unimplemented: Read as '0' T2OUTPS3:T2OUTPS0: Timer2 Output Postscale Select bits
	0000 = 1:1 Postscale 0001 = 1:2 Postscale
	•
	•
	1111 = 1:16 Postscale
bit 2	TMR2ON: Timer2 On bit
	1 = Timer2 is on
	0 = Timer2 is off
bit 1-0	T2CKPS1:T2CKPS0: Timer2 Clock Prescale Select bits
	00 = Prescaler is 1
	01 = Prescaler is 4
	1x = Prescaler is 16

12.2 Timer2 Interrupt

Timer2 can also generate an optional device interrupt. The Timer2 output signal (TMR2 to PR2 match) provides the input for the 4-bit output counter/postscaler. This counter generates the TMR2 match interrupt flag which is latched in TMR2IF (PIR1<1>). The interrupt is enabled by setting the TMR2 Match Interrupt Enable bit, TMR2IE (PIE1<1>).

A range of 16 postscale options (from 1:1 through 1:16 inclusive) can be selected with the postscaler control bits, T2OUTPS3:T2OUTPS0 (T2CON<6:3>).

12.3 Timer2 Output

The unscaled output of TMR2 is available primarily to the CCP modules, where it is used as a time base for operations in PWM mode.

Timer2 can be optionally used as the shift clock source for the MSSP module operating in SPI mode. Additional information is provided in Section 16.0 "Master Synchronous Serial Port (MSSP) Module".

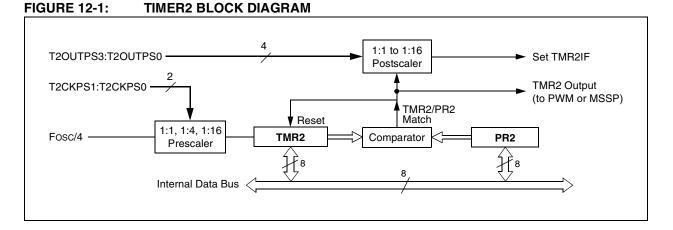


TABLE 12-1: REGISTERS ASSOCIATED WITH TIMER2 AS A TIMER/COUNTER

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INTOIE	RBIE	TMR0IF	INT0IF	RBIF	51
PIR1	—	ADIF	RC1IF	TX1IF	SSPIF	—	TMR2IF	TMR1IF	54
PIE1	—	ADIE	RC1IE	TX1IE	SSPIE	—	TMR2IE	TMR1IE	54
IPR1	—	ADIP	RC1IP	TX1IP	SSPIP	—	TMR2IP	TMR1IP	54
TMR2	Timer2 Register								
T2CON	_	T2OUTPS3	T2OUTPS2	T2OUTPS1	T2OUTPS0	TMR2ON	T2CKPS1	T2CKPS0	52
PR2	Timer2 Peri	iod Register							52

Legend: — = unimplemented, read as '0'. Shaded cells are not used by the Timer2 module.

13.0 TIMER3 MODULE

The Timer3 timer/counter module incorporates these features:

- Software selectable operation as a 16-bit timer or counter
- Readable and writable 8-bit registers (TMR3H and TMR3L)
- Selectable clock source (internal or external) with device clock or Timer1 oscillator internal options
- Interrupt on overflow
- Module Reset on CCP Special Event Trigger

A simplified block diagram of the Timer3 module is shown in Figure 13-1. A block diagram of the module's operation in Read/Write mode is shown in Figure 13-2.

The Timer3 module is controlled through the T3CON register (Register 13-1). It also selects the clock source options for the CCP modules. See **Section 14.2.2** "**Timer1/Timer3 Mode Selection**" for more information.

REGISTER 13-1: T3CON: TIMER3 CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
RD16	T3CCP2	T3CKPS1	T3CKPS0	T3CCP1	T3SYNC	TMR3CS	TMR3ON
bit 7							bit 0

Legend:				
R = Readab	ole bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value a	at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown
bit 7		B-Bit Read/Write Mode Enable		
		les register read/write of Time les register read/write of Time		
bit 6,3	T3CCP2:	T3CCP1: Timer3 and Timer1	to CCPx Enable bits	
	01 = Tim Tim	er3 is the capture/compare cl er1 is the capture/compare cl		
bit 5-4	T3CKPS	1:T3CKPS0: Timer3 Input Cl	ock Prescale Select bits	
	10 = 1:4 01 = 1:2	Prescale value Prescale value Prescale value Prescale value		
bit 2	(Not usat <u>When TM</u> 1 = Do no 0 = Sync <u>When TM</u>	Timer3 External Clock Input ole if the device clock comes IR3CS = 1: ot synchronize external clock hronize external clock input IR3CS = 0: signored. Timer3 uses the int	from Timer1/Timer3.)	0.
bit 1	TMR3CS	: Timer3 Clock Source Selec	t bit	
	fallin	rnal clock input from Timer1 o g edge) nal clock (Fosc/4)	oscillator or T13CKI (on the ris	sing edge after the first
bit 0	TMR3ON	I: Timer3 On bit		
	1 = Enab 0 = Stops	les Timer3 s Timer3		

13.1 Timer3 Operation

Timer3 can operate in one of three modes:

- Timer
- Synchronous Counter
- Asynchronous Counter

The operating mode is determined by the clock select bit, TMR3CS (T3CON<1>). When TMR3CS is cleared (= 0), Timer3 increments on every internal instruction cycle (Fosc/4). When the bit is set, Timer3 increments on every rising edge of the Timer1 external clock input or the Timer1 oscillator, if enabled.

As with Timer1, the RC1/T1OSI and RC0/T1OSO/T13CKI pins become inputs when the Timer1 oscillator is enabled. This means the values of TRISC<1:0> are ignored and the pins are read as '0'.

FIGURE 13-1: TIMER3 BLOCK DIAGRAM (8-BIT MODE)

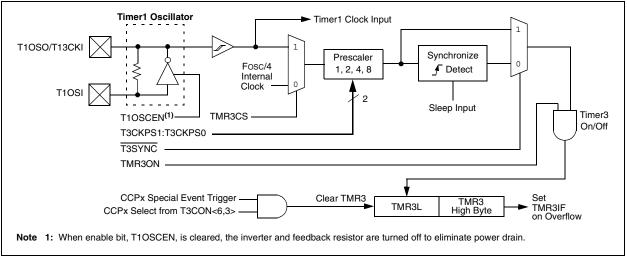
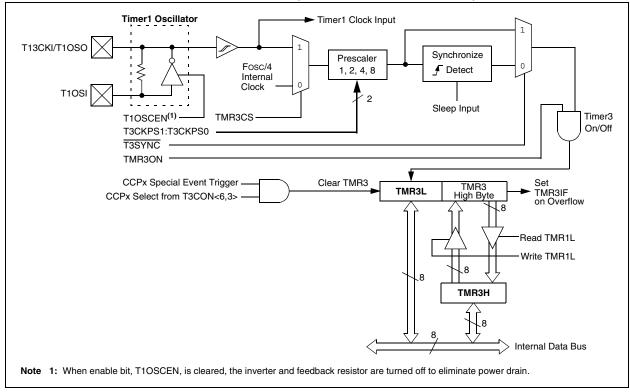


FIGURE 13-2: TIMER3 BLOCK DIAGRAM (16-BIT READ/WRITE MODE)



13.2 Timer3 16-Bit Read/Write Mode

Timer3 can be configured for 16-bit reads and writes (see Figure 13-2). When the RD16 control bit (T3CON<7>) is set, the address for TMR3H is mapped to a buffer register for the high byte of Timer3. A read from TMR3L will load the contents of the high byte of Timer3 into the Timer3 High Byte Buffer register. This provides the user with the ability to accurately read all 16 bits of Timer1 without having to determine whether a read of the high byte, followed by a read of the low byte, has become invalid due to a rollover between reads.

A write to the high byte of Timer3 must also take place through the TMR3H Buffer register. The Timer3 high byte is updated with the contents of TMR3H when a write occurs to TMR3L. This allows a user to write all 16 bits to both the high and low bytes of Timer3 at once.

The high byte of Timer3 is not directly readable or writable in this mode. All reads and writes must take place through the Timer3 High Byte Buffer register.

Writes to TMR3H do not clear the Timer3 prescaler. The prescaler is only cleared on writes to TMR3L.

13.3 Using the Timer1 Oscillator as the Timer3 Clock Source

The Timer1 internal oscillator may be used as the clock source for Timer3. The Timer1 oscillator is enabled by setting the T1OSCEN (T1CON<3>) bit. To use it as the Timer3 clock source, the TMR3CS bit must also be set. As previously noted, this also configures Timer3 to increment on every rising edge of the oscillator source.

The Timer1 oscillator is described in Section 11.0 "Timer1 Module".

13.4 Timer3 Interrupt

The TMR3 register pair (TMR3H:TMR3L) increments from 0000h to FFFFh and overflows to 0000h. The Timer3 interrupt, if enabled, is generated on overflow and is latched in interrupt flag bit, TMR3IF (PIR2<1>). This interrupt can be enabled or disabled by setting or clearing the Timer3 Interrupt Enable bit, TMR3IE (PIE2<1>).

13.5 Resetting Timer3 Using the CCP Special Event Trigger

If CCP1 or CCP2 is configured to use Timer3 and to generate a Special Event Trigger in Compare mode (CCPxM3:CCPxM0 = 1011), this signal will reset Timer3. The trigger from CCP2 will also start an A/D conversion if the A/D module is enabled (see **Section 14.3.4 "Special Event Trigger"** for more information).

The module must be configured as either a timer or synchronous counter to take advantage of this feature. When used this way, the CCPRxH:CCPRxL register pair effectively becomes a period register for Timer3.

If Timer3 is running in Asynchronous Counter mode, the Reset operation may not work.

In the event that a write to Timer3 coincides with a Special Event Trigger from a CCP module, the write will take precedence.

Note: The Special Event Triggers from the CCPx module will not set the TMR3IF interrupt flag bit (PIR2<1>).

TABLE 13-1. REGISTERS ASSOCIATED WITH TIMERS AS A TIMER/COUNTER											
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page		
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51		
PIR2	OSCFIF	CMIF	_	_	BCLIF	LVDIF	TMR3IF	—	54		
PIE2	OSCFIE	CMIE	—	—	BCLIE	LVDIE	TMR3IE	—	54		
IPR2	OSCFIP	CMIP	_	_	BCLIP	LVDIP	TMR3IP	—	54		
TMR3L	Timer3 Reg	gister Low B	yte						53		
TMR3H	Timer3 Register High Byte							53			
T1CON	RD16	T1RUN	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR1ON	52		
T3CON	RD16	T3CCP2	T3CKPS1	T3CKPS0	T3CCP1	T3SYNC	TMR3CS	TMR3ON	53		

 TABLE 13-1:
 REGISTERS ASSOCIATED WITH TIMER3 AS A TIMER/COUNTER

Legend: — = unimplemented, read as '0'. Shaded cells are not used by the Timer3 module.

NOTES:

14.0 CAPTURE/COMPARE/PWM (CCP) MODULES

PIC18F85J90 family devices have two CCP (Capture/Compare/PWM) modules, designated CCP1 and CCP2. Both modules implement standard Capture, Compare and Pulse-Width Modulation (PWM) modes.

Each CCP module contains a 16-bit register which can operate as a 16-bit Capture register, a 16-bit Compare register or a PWM Master/Slave Duty Cycle register. For the sake of clarity, all CCP module operation in the following sections is described with respect to CCP2, but is equally applicable to CCP1.

REGISTER 14-1: CCPxCON: CCPx CONTROL REGISTER (CCP1, CCP2 MODULES)

U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	—	DCxB1	DCxB0	CCPxM3	CCPxM2	CCPxM1	CCPxM0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6 Unimplemented: Read as '0' bit 5-4 DCxB1:DCxB0: PWM Duty Cycle bit 1 and bit 0 for CCPx Module Capture mode: Unused. Compare mode: Unused. PWM mode: These bits are the two Least Significant bits (bit 1 and bit 0) of the 10-bit PWM duty cycle. The eight Most Significant bits (DCx9:DCx2) of the duty cycle are found in CCPRxL. bit 3-0 CCPxM3:CCPxM0: CCPx Module Mode Select bits 0000 = Capture/Compare/PWM disabled (resets CCPx module) 0001 = Reserved 0010 = Compare mode, toggle output on match (CCPxIF bit is set) 0011 = Reserved 0100 = Capture mode, every falling edge 0101 = Capture mode, every rising edge 0110 = Capture mode, every 4th rising edge 0111 = Capture mode, every 16th rising edge 1000 = Compare mode: initialize CCPx pin low; on compare match, force CCPx pin high (CCPxIF bit is set) 1001 = Compare mode: initialize CCPx pin high; on compare match, force CCPx pin low (CCPxIF bit is set) 1010 = Compare mode: generate software interrupt on compare match (CCPxIF bit is set, CCPx pin reflects I/O state) 1011 = Compare mode: Special Event Trigger; reset timer; start A/D conversion on CCPx match (CCPxIF bit is set)⁽¹⁾

- 11xx = PWM mode
- **Note 1:** CCPxM3:CCPxM0 = 1011 will only reset timer and not start A/D conversion on CCP1 match.

14.1 CCP Module Configuration

Each Capture/Compare/PWM module is associated with a control register (generically, CCPxCON) and a data register (CCPRx). The data register in turn is comprised of two 8-bit registers: CCPRxL (low byte) and CCPRxH (high byte). All registers are both readable and writable.

14.1.1 CCP MODULES AND TIMER RESOURCES

The CCP modules utilize Timers 1, 2 or 3, depending on the mode selected. Timer1 and Timer3 are available to modules in Capture or Compare modes, while Timer2 is available for modules in PWM mode.

TABLE 14-1: CCP MODE – TIMER RESOURCE

CCP Mode	Timer Resource			
Capture	Timer1 or Timer3			
Compare	Timer1 or Timer3			
PWM	Timer2			

The assignment of a particular timer to a module is determined by the Timer to CCP enable bits in the T3CON register (Register 13-1). Both modules may be active at any given time and may share the same timer resource if they are configured to operate in the same mode (Capture/Compare or PWM) at the same time. The interactions between the two modules are summarized in Table 14-2.

Depending on the configuration selected, up to four timers may be active at once, with modules in the same configuration (Capture/Compare or PWM) sharing timer resources. The possible configurations are shown in Figure 14-1.

14.1.2 OPEN-DRAIN OUTPUT OPTION

When operating in Output mode (i.e., in Compare or PWM modes), the drivers for the CCPx pins can be optionally configured as open-drain outputs. This feature allows the voltage level on the pin to be pulled to a higher level through an external pull-up resistor and allows the output to communicate with external circuits without the need for additional level shifters.

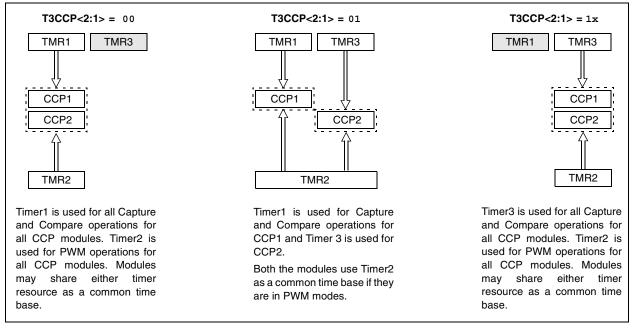
The open-drain output option is controlled by the CCP2OD and CCP1OD bits (TRISG<6:5>). Setting the appropriate bit configures the pin for the corresponding module for open-drain operation.

14.1.3 CCP2 PIN ASSIGNMENT

The pin assignment for CCP2 (Capture input, Compare and PWM output) can change, based on device configuration. The CCP2MX Configuration bit determines which pin CCP2 is multiplexed to. By default, it is assigned to RC1 (CCP2MX = 1). If the Configuration bit is cleared, CCP2 is multiplexed with RE7.

Changing the pin assignment of CCP2 does not automatically change any requirements for configuring the port pin. Users must always verify that the appropriate TRIS register is configured correctly for CCP2 operation, regardless of where it is located.

FIGURE 14-1: CCP AND TIMER INTERCONNECT CONFIGURATIONS



CCP1 Mode	CCP2 Mode	Interaction
Capture	Capture	Each module can use TMR1 or TMR3 as the time base. The time base can be different for each CCP.
Capture	Compare	CCP2 can be configured for the Special Event Trigger to reset TMR1 or TMR3 (depending upon which time base is used). Automatic A/D conversions on trigger event can also be done. Operation of CCP1 could be affected if it is using the same timer as a time base.
Compare	Capture	CCP1 can be configured for the Special Event Trigger to reset TMR1 or TMR3 (depending upon which time base is used). Operation of CCP2 could be affected if it is using the same timer as a time base.
Compare	Compare	Either module can be configured for the Special Event Trigger to reset the time base. Automatic A/D conversions on CCP2 trigger event can be done. Conflicts may occur if both modules are using the same time base.
Capture	PWM	None
Compare	PWM	None
PWM	Capture	None
PWM	Compare	None
PWM	PWM	Both PWMs will have the same frequency and update rate (TMR2 interrupt).

TABLE 14-2: INTERACTIONS BETWEEN CCP1 AND CCP2 FOR TIMER RESOURCES

14.2 Capture Mode

In Capture mode, the CCPR2H:CCPR2L register pair captures the 16-bit value of the TMR1 or TMR3 register when an event occurs on the CCP2 pin (RC1 or RE7, depending on device configuration). An event is defined as one of the following:

- every falling edge
- every rising edge
- · every 4th rising edge
- · every 16th rising edge

The event is selected by the mode select bits, CCP2M3:CCP2M0 (CCP2CON<3:0>). When a capture is made, the interrupt request flag bit, CCP2IF (PIR3<2>), is set; it must be cleared in software. If another capture occurs before the value in register CCPR2 is read, the old captured value is overwritten by the new captured value.

14.2.1 CCP PIN CONFIGURATION

In Capture mode, the appropriate CCPx pin should be configured as an input by setting the corresponding TRIS direction bit.

```
Note: If RC1/CCP2 or RE7/CCP2 is configured 
as an output, a write to the port can cause 
a capture condition.
```

14.2.2 TIMER1/TIMER3 MODE SELECTION

The timers that are to be used with the capture feature (Timer1 and/or Timer3) must be running in Timer mode or Synchronized Counter mode. In Asynchronous Counter mode, the capture operation may not work. The timer to be used with each CCP module is selected in the T3CON register (see Section 14.1.1 "CCP Modules and Timer Resources").

14.2.3 SOFTWARE INTERRUPT

When the Capture mode is changed, a false capture interrupt may be generated. The user should keep the CCP2IE bit (PIE3<2>) clear to avoid false interrupts and should clear the flag bit, CCP2IF, following any such change in operating mode.

14.2.4 CCP PRESCALER

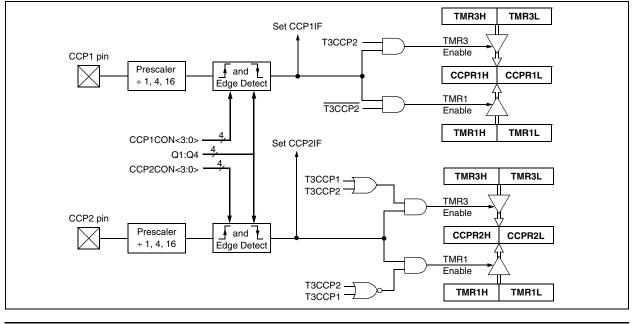
There are four prescaler settings in Capture mode. They are specified as part of the operating mode selected by the mode select bits (CCP2M3:CCP2M0). Whenever the CCP module is turned off, or the CCP module is not in Capture mode, the prescaler counter is cleared. This means that any Reset will clear the prescaler counter.

Switching from one capture prescaler to another may generate an interrupt. Also, the prescaler counter will not be cleared; therefore, the first capture may be from a non-zero prescaler. Example 14-1 shows the recommended method for switching between capture prescalers. This example also clears the prescaler counter and will not generate the "false" interrupt.

EXAMPLE 14-1: CHANGING BETWEEN CAPTURE PRESCALERS

; Turn CCP module off
; Load WREG with the
; new prescaler mode
; value and CCP ON
; Load CCP2CON with
; this value

FIGURE 14-2: CAPTURE MODE OPERATION BLOCK DIAGRAM



14.3 Compare Mode

In Compare mode, the 16-bit CCPR2 register value is constantly compared against either the TMR1 or TMR3 register pair value. When a match occurs, the CCP2 pin can be:

- driven high
- driven low
- toggled (high-to-low or low-to-high)
- remain unchanged (that is, reflects the state of the I/O latch)

The action on the pin is based on the value of the mode select bits (CCP2M3:CCP2M0). At the same time, the interrupt flag bit, CCP2IF, is set.

14.3.1 CCP PIN CONFIGURATION

The user must configure the CCPx pin as an output by clearing the appropriate TRIS bit.

Clearing the CCP2CON register will force					
the RC1 or RE7 compare output latch					
(depending on device configuration) to the					
default low level. This is not the PORTC or					
PORTE I/O data latch.					

14.3.2 TIMER1/TIMER3 MODE SELECTION

Timer1 and/or Timer3 must be running in Timer mode, or Synchronized Counter mode, if the CCP module is using the compare feature. In Asynchronous Counter mode, the compare operation may not work.

14.3.3 SOFTWARE INTERRUPT MODE

When the Generate Software Interrupt mode is chosen (CCP2M3:CCP2M0 = 1010), the CCP2 pin is not affected. Only a CCP interrupt is generated, if enabled, and the CCP2IE bit is set.

14.3.4 SPECIAL EVENT TRIGGER

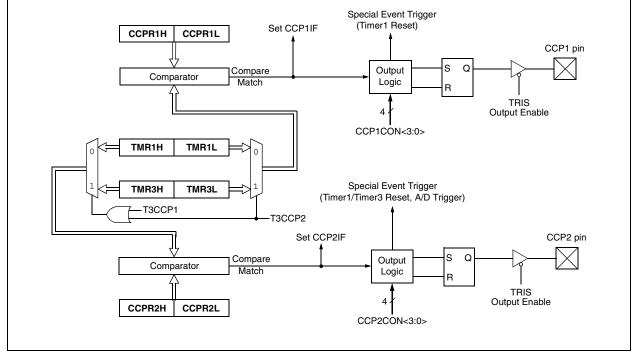
Both CCP modules are equipped with a Special Event Trigger. This is an internal hardware signal generated in Compare mode to trigger actions by other modules. The Special Event Trigger is enabled by selecting the Compare Special Event Trigger mode (CCP2M3:CCP2M0 = 1011).

For either CCP module, the Special Event Trigger resets the timer register pair for whichever timer resource is currently assigned as the module's time base. This allows the CCPRx registers to serve as a programmable period register for either timer.

The Special Event Trigger for CCP2 can also start an A/D conversion. In order to do this, the A/D converter must already be enabled.

Note: The Special Event Trigger of CCP1 only resets Timer1/Timer3 and cannot start an A/D conversion even when the A/D converter is enabled.

FIGURE 14-3: COMPARE MODE OPERATION BLOCK DIAGRAM



Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51
RCON	IPEN	_		RI	TO	PD	POR	BOR	52
PIR3	—	LCDIF	RC2IF	TX2IF	—	CCP2IF	CCP1IF	_	54
PIE3	—	LCDIE	RC2IE	TX2IE	—	CCP2IE	CCP1IE	—	54
IPR3	—	LCDIP	RC2IP	TX2IP	—	CCP2IP	CCP1IP	_	54
PIR2	OSCFIF	CMIF		—	BCLIF	LVDIF	TMR3IF	_	54
PIE2	OSCFIE	CMIE	_	_	BCLIE	LVDIE	TMR3IE	_	54
IPR2	OSCFIP	CMIP	_	_	BCLIP	LVDIP	TMR3IP	_	54
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	54
TRISE	TRISE7	TRISE6	TRISE5	TRISE4	TRISE3	_	TRISE1	TRISE0	54
TRISG	SPIOD	CCP2OD	CCP10D	TRISG4	TRISG3	TRISG2	TRISG1	TRISG0	54
TMR1L	Timer1 Reg	gister Low B	yte						52
TMR1H	Timer1 Reg	gister High E	Byte						52
T1CON	RD16	T1RUN	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N	52
TMR3H	Timer3 Reg	gister High E	Byte						53
TMR3L	Timer3 Reg	gister Low B	yte						53
T3CON	RD16	T3CCP2	T3CKPS1	T3CKPS0	T3CCP1	T3SYNC	TMR3CS	TMR3ON	53
CCPR1L	Capture/Co	ompare/PWI	M Register	Low Byte					55
CCPR1H	Capture/Co	ompare/PWI	M Register	I High Byte					55
CCP1CON	—	—	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	55
CCPR2L	Capture/Co	ompare/PWI	M Register 2	2 Low Byte					56
CCPR2H	Capture/Co	ompare/PWI	A Register 2	2 High Byte					55
CCP2CON	—	—	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	56

TABLE 14-3: REGISTERS ASSOCIATED WITH CAPTURE, COMPARE, TIMER1 AND TIMER3

Legend: — = unimplemented, read as '0'. Shaded cells are not used by Capture/Compare, Timer1 or Timer3.

14.4 PWM Mode

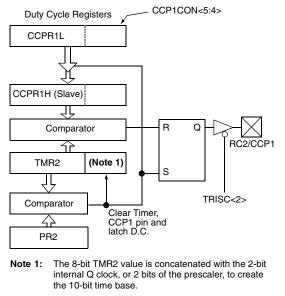
In Pulse-Width Modulation (PWM) mode, the CCP2 pin produces up to a 10-bit resolution PWM output. Since the CCP2 pin is multiplexed with a PORTC or PORTE data latch, the appropriate TRIS bit must be cleared to make the CCP2 pin an output.

Note:	Clearing the CCP2CON register will force the RC1 or RE7 output latch (depending on device configuration) to the default low level. This is not the PORTC or PORTE
	I/O data latch.

Figure 14-4 shows a simplified block diagram of the CCP1 module in PWM mode.

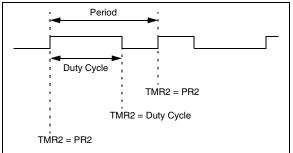
For a step-by-step procedure on how to set up the CCP module for PWM operation, see **Section 14.4.3** "Setup for PWM Operation".





A PWM output (Figure 14-5) has a time base (period) and a time that the output stays high (duty cycle). The frequency of the PWM is the inverse of the period (1/period).

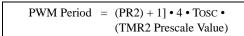
FIGURE 14-5: PWM OUTPUT



14.4.1 PWM PERIOD

The PWM period is specified by writing to the PR2 register. The PWM period can be calculated using the following formula:

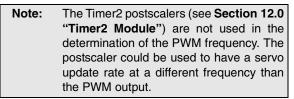
EQUATION 14-1:



PWM frequency is defined as 1/[PWM period].

When TMR2 is equal to PR2, the following three events occur on the next increment cycle:

- TMR2 is cleared
- The CCP2 pin is set (exception: if PWM duty cycle = 0%, the CCP2 pin will not be set)
- The PWM duty cycle is latched from CCPR2L into CCPR2H



14.4.2 PWM DUTY CYCLE

The PWM duty cycle is specified by writing to the CCPR2L register and to the CCP2CON<5:4> bits. Up to 10-bit resolution is available. The CCPR2L contains the eight MSbs and the CCP2CON<5:4> contains the two LSbs. This 10-bit value is represented by CCPR2L:CCP2CON<5:4>. The following equation is used to calculate the PWM duty cycle in time:

EQUATION 14-2:

PWM Duty Cycle = (CCPR2L:CCP2CON<5:4>) • Tosc • (TMR2 Prescale Value)

CCPR2L and CCP2CON<5:4> can be written to at any time, but the duty cycle value is not latched into CCPR2H until after a match between PR2 and TMR2 occurs (i.e., the period is complete). In PWM mode, CCPR2H is a read-only register.

The CCPR2H register and a 2-bit internal latch are used to double-buffer the PWM duty cycle. This double-buffering is essential for glitchless PWM operation.

When the CCPR2H and 2-bit latch match TMR2, concatenated with an internal 2-bit Q clock or 2 bits of the TMR2 prescaler, the CCP2 pin is cleared.

The maximum PWM resolution (bits) for a given PWM frequency is given by the equation:

EQUATION 14-3:

PWM Resolution (max) =
$$\frac{\log(\frac{FOSC}{FPWM})}{\log(2)}$$
 bits

Note: If the PWM duty cycle value is longer than the PWM period, the CCP2 pin will not be cleared.

TABLE 14-4: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS AT 40 MHz

PWM Frequency	2.44 kHz	9.77 kHz	39.06 kHz	156.25 kHz	312.50 kHz	416.67 kHz
Timer Prescaler (1, 4, 16)	16	4	1	1	1	1
PR2 Value	FFh	FFh	FFh	3Fh	1Fh	17h
Maximum Resolution (bits)	14	12	10	8	7	6.58

14.4.3 SETUP FOR PWM OPERATION

The following steps should be taken when configuring the CCP module for PWM operation:

- 1. Set the PWM period by writing to the PR2 register.
- 2. Set the PWM duty cycle by writing to the CCPR2L register and CCP2CON<5:4> bits.
- 3. Make the CCP2 pin an output by clearing the appropriate TRIS bit.
- 4. Set the TMR2 prescale value, then enable Timer2 by writing to T2CON.
- 5. Configure the CCP2 module for PWM operation.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51
RCON	IPEN	_	_	RI	TO	PD	POR	BOR	52
PIR1	_	ADIF	RC1IF	TX1IF	SSPIF	—	TMR2IF	TMR1IF	54
PIE1	_	ADIE	RC1IE	TX1IE	SSPIE		TMR2IE	TMR1IE	54
IPR1	_	ADIP	RC1IP	TX1IP	SSPIP		TMR2IP	TMR1IP	54
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	54
TRISE	TRISE7	TRISE6	TRISE5	TRISE4	TRISE3		TRISE1	TRISE0	54
TRISG	SPIOD	CCP2OD	CCP10D	TRISG4	TRISG3	TRISG2	TRISG1	TRISG0	54
TMR2	Timer2 Reg	gister							52
PR2	Timer2 Per	iod Register							52
T2CON		T2OUTPS3	T2OUTPS2	T2OUTPS1	T2OUTPS0	TMR2ON	T2CKPS1	T2CKPS0	52
CCPR1L	Capture/Co	mpare/PWN	I Register 1 I	_ow Byte					55
CCPR1H	Capture/Co	mpare/PWN	I Register 1 I	High Byte					55
CCP1CON		_	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	55
CCPR2L	Capture/Co	mpare/PWN	I Register 2 I	_ow Byte					56
CCPR2H	Capture/Compare/PWM Register 2 High Byte						55		
CCP2CON			DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	56

TABLE 14-5: REGISTERS ASSOCIATED WITH PWM AND TIMER2

Legend: — = unimplemented, read as '0'. Shaded cells are not used by PWM or Timer2.

NOTES:

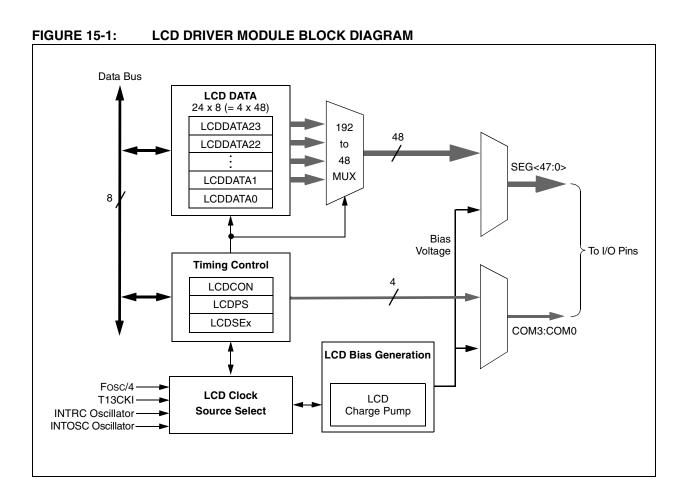
15.0 LIQUID CRYSTAL DISPLAY (LCD) DRIVER MODULE

The Liquid Crystal Display (LCD) driver module generates the timing control to drive a static or multiplexed LCD panel. It also provides control of the LCD pixel data. The module can drive panels of up to 192 pixels (48 segments by 4 commons) in 80-pin devices, and 132 pixels (33 segments by 4 commons) in 64-pin devices.

The LCD driver module supports these features:

- Direct driving of LCD panel
- On-chip bias generator with dedicated charge pump to support a range of fixed and variable bias options
- Up to four commons, with four Multiplexing modes
- Up to 48 (80-pin devices) or 33 (64-pin devices) segments
- Three LCD clock sources with selectable prescaler, with a fourth source available for use with the LCD charge pump

A simplified block diagram of the module is shown in Figure 15-1.



15.1 LCD Registers

The LCD driver module has 33 registers:

- LCD Control Register (LCDCON)
- LCD Phase Register (LCDPS)
- LCDREG Register (LCD Regulator Control)
- Six LCD Segment Enable Registers (LCDSE5:LCDSE0)
- 24 LCD Data Registers (LCDDATA23:LCDDATA0)

15.1.1 LCD CONTROL REGISTERS

The LCDCON register, shown in Register 15-1, controls the overall operation of the module. Once the module is configured, the LCDEN (LCDCON<7>) bit is used to enable or disable the LCD module. The LCD panel can also operate during Sleep by clearing the SLPEN (LCDCON<6>) bit.

The LCDPS register, shown in Register 15-2, configures the LCD clock source prescaler and the type of waveform: Type-A or Type-B. Details on these features are provided in Section 15.2 "LCD Clock Source", Section 15.3 "LCD Bias Generation" and Section 15.8 "LCD Waveform Generation".

The LCDREG register is described in Section 15.3 "LCD Bias Generation".

The LCD Segment Enable registers (LCDSEx) configure the functions of the port pins. Setting the segment enable bit for a particular segment configures that pin as an LCD driver. The prototype LCDSE register is shown in Register 15-3. There are six LCDSE registers (LCDSE5:LCDSE0), listed in Table 15-1.

REGISTER 15-1: LCDCON: LCD CONTROL REGISTER

R/W-0	R/W-0	R/C-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0
LCDEN	SLPEN	WERR	—	CS1	CS0	LMUX1	LMUX0
bit 7							bit (
Legend:		C = Clearable	bit				
R = Readat	ble bit	W = Writable	bit	U = Unimple	mented bit, read	as '0'	
-n = Value a	at POR	'1' = Bit is set		'0' = Bit is cl	eared	x = Bit is unkr	nown
bit 7	1 = LCD driv	D Driver Enable ver module is en ver module is dis	abled				
bit 6	1 = LCD driv	D Driver Enable ver module is dis ver module is en	abled in Sleep	mode			
bit 5		•		PS<4> = 0 (mi	ust be cleared in	software)	
bit 4	Unimpleme	nted: Read as 'o	כי				
bit 3-2	1x = INTRC 01 = T13CK	. ,	ect bits				
bit 1-0	-	JX0: Commons	Select bits				
	LMUX1:	NA	-	Maximum Nun	nber of Pixels:	D!	Turne
	LMUX0	Multiplex 1		C18F6XJ90	PIC18F8XJ90	– Bias	гуре
						-	

00

01

10

11

Static (COM0)

1/2 (COM1:COM0)

1/3 (COM2:COM0)

1/4 (COM3:COM0)

33

66

99

132

48

96

144

192

Static

1/2 or 1/3

1/2 or 1/3

1/3

R/W-0	R/W-0	R-0	R-0	R/W-0	R/W-0	R/W-0	R/W-0
WFT	BIASMD	LCDA	WA	LP3	LP2	LP1	LP0
bit 7							bit (
Levende							
Legend: R = Readab	le hit	W = Writable I	nit	LI = Unimplen	nented bit, rea	d as '0'	
-n = Value at		'1' = Bit is set		'0' = Bit is clea		x = Bit is unkn	lown
bit 7	WFT: Wavefo	rm Type Select	bit				
				each frame bou			
h it 0			-	hin each commo	on type)		
bit 6		s Mode Select :LMUX0 = 00:	DIL				
		s mode (do not	set this bit to	· '1')			
		:LMUX0 = 01 0		,			
	1 = 1/2 Bias n						
	0 = 1/3 Bias n						
		:LMUX0 = 11: node (do not se	et this bit to '1	')			
bit 5		ctive Status bit		,			
		r module is act					
		r module is ina					
bit 4	-	te Allow Status					
		the LCDDATA					
bit 3-0		D Prescaler Se	-				
	1111 = 1:16						
	1110 = 1:15						
	1101 = 1:14 1100 = 1:13						
	1000 = 1.13 1011 = 1:12						
	1010 = 1:11						
	1001 = 1:10						
	1000 = 1:9 0111 = 1:8						
	0110 = 1.3						
	0101 = 1 :6						
	0100 = 1:5						
	0011 = 1:4 0010 = 1:3						
	0001 = 1:2						
	0000 = 1:1						

REGISTER 15-2: LCDPS: LCD PHASE REGISTER

REGISTER 15-3: LCDSEx: LCD SEGMENT ENABLE REGISTERS

R/W-0	R/W-0						
SE(n + 7)	SE(n + 6)	SE(n + 5)	SE(n + 4)	SE(n + 3)	SE(n + 2)	SE(n + 1)	SE(n)
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	t, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 SEG(n + 7):SEG(n): Segment Enable bits

For LCDSE0: n = 0For LCDSE1: n = 8For LCDSE2: n = 16For LCDSE3: n = 24For LCDSE4: n = 32For LCDSE5: n = 401 = Segment function of the pin is enabled, digital I/O disabled

0 = I/O function of the pin is enabled

TABLE 15-1: LCDSE REGISTERS AND ASSOCIATED SEGMENTS

Register	Segments
LCDSE0	7:0
LCDSE1	15:8
LCDSE2	23:16
LCDSE3	31:24
LCDSE4 ⁽¹⁾	39:32
LCDSE5 ⁽²⁾	47:40

Note 1: LCDSE4<7:1> (SEG39:SEG33) are not implemented in 64-pin devices.

2: LCDSE5 is not implemented in 64-pin devices.

15.1.2 LCD DATA REGISTERS

Once the module is initialized for the LCD panel, the individual bits of the LCDDATA23:LCDDATA0 registers are cleared or set to represent a clear or dark pixel, respectively. Specific sets of LCDDATA registers are used with specific segments and common signals. Each bit represents a unique combination of a specific segment connected to a specific common.

Individual LCDDATA bits are named by the convention "SxxCy", with "xx" as the segment number and "y" as the common number. The relationship is summarized in Table 15-2. The prototype LCDDATA register is shown in Register 15-4.

Note: In 64-pin devices, writing into the registers LCDDATA5, LCDDATA11, LCDDATA17, and LCDDATA23 will not affect the status of any pixels.

REGISTER 15-4: LCDDATAX: LCD DATA REGISTERS

R/W-0	R/W-0						
S(n + 7)Cy	S(n + 6)Cy	S(n + 5)Cy	S(n + 4)Cy	S(n + 3)Cy	S(n + 2)Cy	S(n + 1)Cy	S(n)Cy
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	1 as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0	S(n + 7)Cy:S(n)Cy: Pixel On bits
	For LCDDATA0 through LCDDATA5: n = (8x), y = 0
	For LCDDATA6 through LCDDATA11: n = (8(x - 6)), y = 1
	<u>For LCDDATA12 through LCDDATA17: n = (8(x – 12)), y = 2</u>
	For LCDDATA18 through LCDDATA23: $n = (8(x - 18)), y = 3$
	1 = Pixel on (dark)
	0 = Pixel off (clear)

TABLE 15-2: LCDDATA REGISTERS AND BITS FOR SEGMENT AND COM COMBINATIONS

Commonto	COM Lines						
Segments	0	1	2	3			
O thursday 7	LCDDATA0	LCDDATA6	LCDDATA12	LCDDATA18			
0 through 7	S00C0:S07C0	S00C1:S07C1	S00C2:S07C2	S00C3:S07C3			
8 through 15	LCDDATA1	LCDDATA7	LCDDATA13	LCDDATA19			
	S08C0:S15C0	S08C1:S15C1	S08C2:S15C2	S08C0:S15C3			
16 through 23	LCDDATA2	LCDDATA8	LCDDATA14	LCDDATA20			
16 through 23	S16C0:S23C0	S16C1:S23C1	S16C2:S23C2	S16C3:S23C3			
24 through 31	LCDDATA3	LCDDATA9	LCDDATA15	LCDDATA21			
24 (mough 51	S24C0:S31C0	S24C1:S31C1	S24C2:S31C2	S24C3:S31C3			
20 through 20	LCDDATA4 ⁽¹⁾	LCDDATA10 ⁽¹⁾	LCDDATA16 ⁽¹⁾	LCDDATA22 ⁽¹⁾			
32 through 39	S32C0:S39C0	S32C1:S39C1	S32C2:S39C2	S32C3:S39C3			
40 through 47	LCDDATA5 ⁽²⁾	LCDDATA11 ⁽²⁾	LCDDATA17 ⁽²⁾	LCDDATA23 ⁽²⁾			
40 iiii0ugii 47	S40C0:S47C0	S40C1:S47C1	S40C2:S47C2	S40C3:S47C3			

Note 1: Bits<7:1> of these registers are not implemented in 64-pin devices. Bit 0 of these registers (SEG32Cy) is always implemented.

2: These registers are not implemented on 64-pin devices.

15.2 LCD Clock Source

The LCD driver module generates its internal clock from 3 possible sources:

- System clock (Fosc/4)
- Timer1 oscillator
- INTRC source

The LCD clock generator uses a configurable divide-by-32/divide-by-8192 postscaler to produce a baseline frequency of about 1 kHz nominal, regardless of the source selected. The clock source selection and the postscaler configuration are determined by the Clock Source Select bits, CS1:CS0 (LCDCON<3:2>).

An additional programmable prescaler is used to derive the LCD frame frequency from the 1 kHz baseline. The prescaler is configured using the LP3:LP0 bits (LCDPS<3:0>) for any one of 16 options, ranging from 1:1 to 1:16.

Proper timing for waveform generation is set by the LMUX1:LMUX0 bits (LCDCON<1:0>). These bits determine which Commons Multiplexing mode is to be used, and divide down the LCD clock source as required. They also determine the configuration of the ring counter that is used to switch the LCD commons on or off.

15.2.1 LCD VOLTAGE REGULATOR CLOCK SOURCE

In addition to the clock source for LCD timing, a separate 31 kHz nominal clock is required for the LCD charge pump. This is provided from a distinct branch of the LCD clock source.

The charge pump clock can use either the Timer1 oscillator or the INTRC source, as well as the 8 MHz INTOSC source (after being divided by 256 by a prescaler). The charge pump clock source is configured using the CKSEL1:CKSEL0 bits (LCDREG<1:0>).

15.2.2 CLOCK SOURCE CONSIDERATIONS

When using the system clock as the LCD clock source, it is assumed that the system clock frequency is a nominal 32 MHz (for a FOSC/4 frequency of 8 MHz). Because the prescaler option for the FOSC/4 clock selection is fixed at divide-by-8192, system clock speeds that differ from 32 MHz will produce frame frequencies and refresh rates different than discussed in this chapter. The user will need to keep this in mind when designing the display application.

The Timer1 and INTRC sources can be used as LCD clock sources when the device is in Sleep mode. To use the Timer1 oscillator, it is necessary to set the T1OSCEN bit (T1CON<3>). Selecting either Timer1 or INTRC as the LCD clock source will not automatically activate these sources.

Similarly, selecting the INTOSC as the charge pump clock source will not turn the oscillator on. To use INTOSC, it must be selected as the system clock source by using the FOSC2 Configuration bit.

If Timer1 is used as a clock source for the device, either as an LCD clock source or for any other purpose, LCD segment 32 become unavailable.

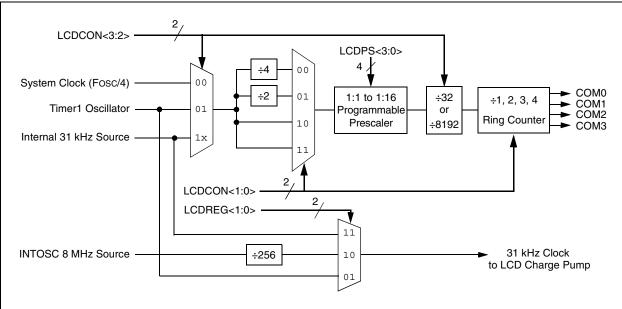


FIGURE 15-2: LCD CLOCK GENERATION

15.3 LCD Bias Generation

The LCD driver module is capable of generating the required bias voltages for LCD operation with a minimum of external components. This includes the ability to generate the different voltage levels required by the different bias types required by the LCD. The driver module can also provide bias voltages both above and below microcontroller VDD through the use of an on-chip LCD voltage regulator.

15.3.1 LCD BIAS TYPES

PIC18F85J90 family devices support three bias types based on the waveforms generated to control segments and commons:

- Static (two discrete levels)
- 1/2 Bias (three discrete levels
- 1/3 Bias (four discrete levels)

The use of different waveforms in driving the LCD is discussed in more detail in **Section 15.8** "LCD Waveform **Generation**".

15.3.2 LCD VOLTAGE REGULATOR

The purpose of the LCD regulator is to provide proper bias voltage and good contrast for the LCD, regardless of VDD levels. This module contains a charge pump and internal voltage reference. The regulator can be configured by using external components to boost bias voltage above VDD. It can also operate a display at a constant voltage below VDD. The regulator can also be selectively disabled to allow bias voltages to be generated by an external resistor network.

The LCD regulator is controlled through the LCDREG register (Register 15-5). It is enabled or disabled using the CKSEL1:CKSEL0 bits, while the charge pump can be selectively enabled using the CPEN bit. When the regulator is enabled, the MODE13 bit is used to select the bias type. The peak LCD bias voltage, measured as a difference between the potentials of LCDBIAS3 and LCDBIAS0, is configured with the BIAS bits.

REGISTER 15-5: LCDREG: VOLTAGE REGULATOR CONTROL REGISTER

U-0	RW-0	RW-1	RW-1	RW-1	RW-1	RW-0	RW-0
—	CPEN	BIAS2	BIAS1	BIAS0	MODE13	CKSEL1	CKSEL0
bit 7					•		bit 0

Legend:									
R = Readab	ole bit	W = Writable bit	U = Unimplemented bit	, read as '0'					
-n = Value a	at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown					
bit 7	Unimpleme	ented: Read as '0'							
bit 6	CPEN: LCE	O Charge Pump Enable bit							
	•	e pump enabled; highest L e pump disabled; highest L	C C						
bit 5-3	BIAS2:BIA	BIAS2:BIAS0: Regulator Voltage Output Control bits							
	110 = 3.47' $101 = 3.34'$ $100 = 3.21'$ $011 = 3.08'$ $010 = 2.95'$ $001 = 2.82'$	V peak (offset on LCDBIAS V peak (offset on LCDBIAS	S0 of 0.13V) S0 of 0.26V) S0 of 0.39V) S0 of 0.52V) S0 of 0.65V) S0 of 0.78V)						
bit 2	1 = Regula	I/3 LCD Bias Enable bit tor output supports 1/3 LC tor output supports static I							
bit 1-0	CKSEL1:C	KSEL0: Regulator Clock S	Source Select bits						

bit 1-0 CKSEL1:CKSEL0: Regulator Clock Source Select bits

- 11 = INTRC
- 10 = INTOSC 8 MHz source
- 01 = Timer1 oscillator
- 00 = LCD regulator disabled

15.3.3 BIAS CONFIGURATIONS

PIC18F85J90 family devices have four distinct circuit configurations for LCD bias generation:

- M0: Regulator with Boost
- M1: Regulator without Boost
- M2: Resistor Ladder with Software Contrast
- M3: Resistor Ladder with Hardware Contrast

15.3.3.1 M0 (Regulator with Boost)

In M0 operation, the LCD charge pump feature is enabled. This allows the regulator to generate voltages up to +3.6V to the LCD (as measured at LCDBIAS3).

M0 uses a flyback capacitor connected between VLCAP1 and VLCAP2, as well as filter capacitors on LCDBIAS0 through LCDBIAS3, to obtain the required voltage boost (Figure 15-3). The output voltage (VBIAS) is the difference of potential between LCDBIAS3 and LCDBIAS0. It is set by the BIAS2:BIAS0 bits which adjust the offset between LCDBIAS0 and Vss. The flyback capacitor (CFLY) acts as a charge storage element for large LCD loads. This mode is useful in those cases where the voltage requirements of the LCD are higher than the microcontroller's VDD. It also permits software control of the display's contrast by adjustment of bias voltage by changing the value of the BIAS bits.

M0 supports Static and 1/3 Bias types. Generation of the voltage levels for 1/3 Bias is handled automatically, but must be configured in software.

M0 is enabled by selecting a valid regulator clock source (CKSEL<1:0> set to any value except '00') and setting the CPEN bit. If static Bias type is required, the MODE13 bit must be cleared.

15.3.3.2 M1 (Regulator without Boost)

M1 operation is similar to M0, but does not use the LCD charge pump. It can provide VBIAS up to the voltage level supplied directly to LCDBIAS3. It can be used in cases where VDD for the application is expected to never drop below a level that can provide adequate contrast for the LCD. The connection of external components is very similar to M0, except that LCDBIAS3 must be tied directly to VDD (Figure 15-3).

The BIAS<2:0> bits can still be used to adjust contrast in software by changing VBIAS. As with M0, changing these bits changes the offset between LCDBIAS0 and Vss. In M1, this is reflected in the change between the LCDBIAS0 and the voltage tied to LCDBIAS3. Thus, if VDD should change, VBIAS will also change; where in M0, the level of VBIAS is constant.

Like M0, M1 supports Static and 1/3 Bias types.Generation of the voltage levels for 1/3 Bias is handled automatically but must be configured in software.

M1 is enabled by selecting a valid regulator clock source (CKSEL<1:0> set to any value except '00') and clearing the CPEN bit. If 1/3 Bias type is required, the MODE13 bit should also be set.

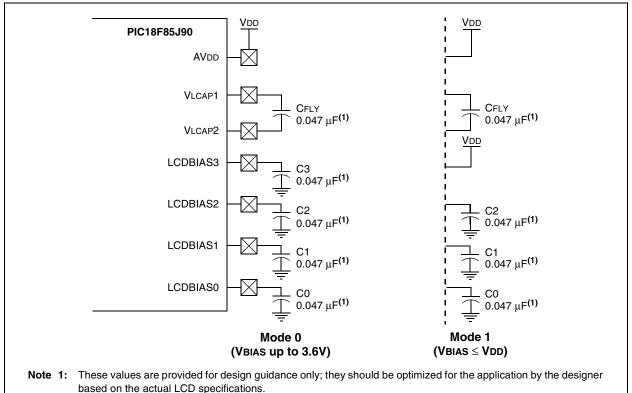


FIGURE 15-3: LCD REGULATOR CONNECTIONS FOR M0 AND M1 CONFIGURATIONS

15.3.3.3 M2 (Resistor Ladder with Software Contrast)

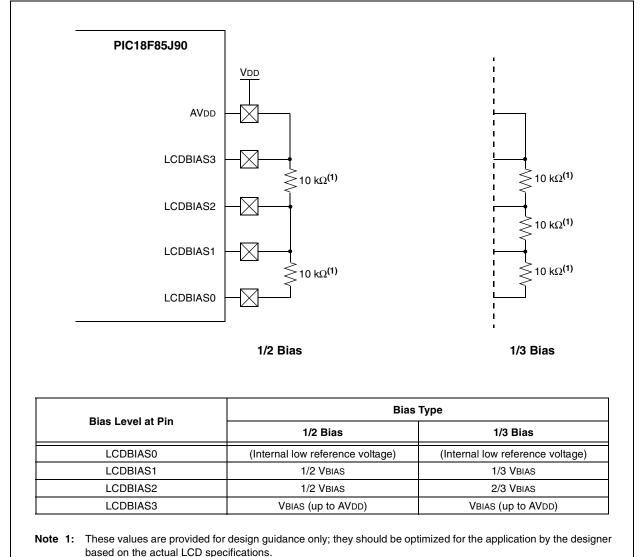
M2 operation also uses the LCD regulator but disables the charge pump. The regulator's internal voltage reference remains active as a way to regulate contrast. It is used in cases where the current requirements of the LCD exceed the capacity of the regulator's charge pump.

In this configuration, the LCD bias voltage levels are created by an external resistor voltage divider connected across LCDBIAS0 through LCDBIAS3, with the top of the divider tied to VDD (Figure 15-4). The potential at the bottom of the ladder is determined by the LCD regulator's voltage reference, tied internally to LCDBIAS0. The bias type is determined by the voltages on the LCDBIAS pins, which are controlled by the configuration of the resistor ladder. Most applications using M2 will use a 1/3 or 1/2 Bias type. While Static Bias can also be used, it offers extremely limited contrast range and additional current consumption over other bias generation modes.

Like M1, the LCDBIAS bits can be used to control contrast, limited by the level of VDD supplied to the device. Also, since there is no capacitor required across VLCAP1 and VLCAP2, these pins are available as digital I/O ports, RG2 and RG3.

M2 is selected by clearing the CKSEL<1:0> bits and setting the CPEN bit.

FIGURE 15-4: RESISTOR LADDER CONNECTIONS FOR CONFIGURATION M2



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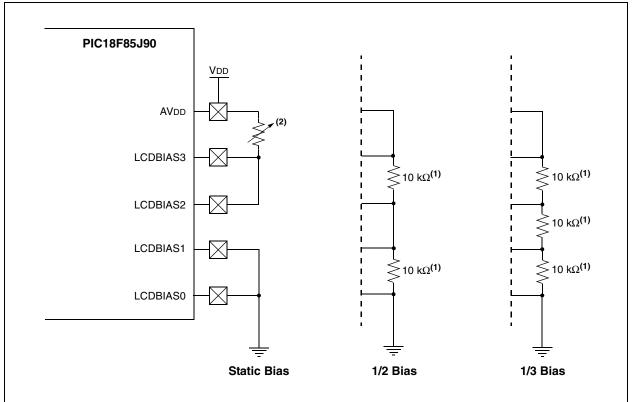
15.3.3.4 M3 (Hardware Contrast)

In M3, the LCD regulator is completely disabled. Like M2, LCD bias levels are tied to AVDD, and are generated using an external divider. The difference is that the internal voltage reference is also disabled and the bottom of the ladder is tied to ground (Vss); see Figure 15-5. The value of the resistors and the difference between Vss and VDD determine the contrast range; no software adjustment is possible. This configuration is also used where the LCD's current requirements exceed the capacity of the charge pump, and software contrast control is not needed.

Depending on the bias type required, resistors are connected between some or all of the pins. A potentiometer can also be connected between LCDBIAS3 and VDD to allow for hardware controlled contrast adjustment.

M3 is selected by clearing the CKSEL<1:0> and CPEN bits.

FIGURE 15-5: RESISTOR LADDER CONNECTIONS FOR CONFIGURATION M3



Bias Level at Pin	Bias Type		
	Static	1/2 Bias	1/3 Bias
LCDBIAS0	AVss	AVss	AVss
LCDBIAS1	AVss	1/2 AVDD	1/3 AVDD
LCDBIAS2	AVDD	1/2 AVDD	2/3 AVDD
LCDBIAS3	AVDD	AVDD	AVDD

Note 1: These values are provided for design guidance only; they should be optimized for the application by the designer based on the actual LCD specifications.

2: Potentiometer for manual contrast adjustment is optional; it may be omitted entirely.

15.3.4 DESIGN CONSIDERATIONS FOR THE LCD CHARGE PUMP

When designing applications that use the LCD regulator with the charge pump enabled, users must always consider both the dynamic current and RMS (static) current requirements of the display, and what the charge pump can deliver. Both dynamic and static current can be determined by Equation 15-1:

EQUATION 15-1:

$$I = C x \frac{dV}{dT}$$

For dynamic current, *C* is the value of the capacitors attached to LCDBIAS3 and LCDBIAS2. The variable, *dV*, is the voltage drop allowed on C2 and C3 during a voltage switch on the LCD display, and *dT* is the duration of the transient current after a clock pulse occurs. For practical design purposes, these will be assumed to be 0.047 μ F for *C*, 0.1V for *dV* and 1 μ s for *dT*. This yields a dynamic current of 4.7 mA for 1 μ s.

RMS current is determined by the value of CFLY for *C*, the voltage across VLCAP1 and VLCAP2 for *dV* and the regulator clock period (TPER) for *dT*. Assuming CFLY of 0.047 μ F, a value of 1.02V across CFLY and TPER of 30 μ s, the maximum theoretical static current will be 1.8 mA. Since the charge pump must charge five capacitors, the maximum current becomes 360 μ A. For a real-world assumption of 50% efficiency, this yields a practical current of 180 μ A.

Users should compare the calculated current capacity against the requirements of the LCD. While dV and dT are relatively fixed by device design, the values of CFLY and the capacitors on the LCDBIAS pins can be changed to increase or decrease current. As always, any changes should be evaluated in the actual circuit for its impact on the application.

15.4 LCD Multiplex Types

The LCD driver module can be configured into four multiplex types:

- Static (only COM0 used)
- 1/2 multiplex (COM0 and COM1 are used)
- 1/3 multiplex (COM0, COM1 and COM2 are used)
- 1/4 multiplex (all COM0, COM1, COM2 and COM3 are used)

The number of active commons used is configured by the LMUX1:LMUX0 bits (LCDCON<1:0>), which determines the function of the PORTE<6:4> pins (see Table 15-3 for details). If the pin is configured as a COM drive, the port I/O function is disabled and the TRIS setting of that pin is overridden.

Note:	On a Power-on Reset, the LMUX1:LMUX0
	bits are '00'.

TABLE 15-3: PORTE<6:4> FUNCTION

LMUX1: LMUX0	PORTE<6>	PORTE<5>	PORTE<4>
00	Digital I/O	Digital I/O	Digital I/O
01	Digital I/O	Digital I/O	COM1 Driver
10	Digital I/O	COM2 Driver	COM1 Driver
11	COM3 Driver	COM2 Driver	COM1 Driver

15.5 Segment Enables

The LCDSEx registers are used to select the pin function for each segment pin. Setting a bit configures the corresponding pin to function as a segment driver. LCDSEx registers do not override the TRIS bit settings, so the TRIS bits must be configured as input for that pin.

Note: On a Power-on Reset, these pins are configured as digital I/O.

15.6 Pixel Control

The LCDDATAx registers contain bits which define the state of each pixel. Each bit defines one unique pixel.

Table 15-2 shows the correlation of each bit in the LCDDATAx registers to the respective common and segment signals. Any LCD pixel location not being used for display can be used as general purpose RAM.

15.7 LCD Frame Frequency

The rate at which the COM and SEG outputs changes is called the LCD frame frequency. Frame frequency is set by the LP3:LP0 bits (LCDPS<3:0>), and is also affected by the Multiplex mode being used. The relationship between the Multiplex mode, LP bits setting and frame rate is shown in Table 15-4 and Table 15-5.

TABLE 15-4:	FRAME FREQUENCY
	FORMULAS

Multiplex Mode	Frame Frequency (Hz)	
Static	Clock source/(4 x 1 x (LP3:LP0 + 1))	
1/2	Clock source/(2 x 2 x (LP3:LP0 + 1))	
1/3	Clock source/(1 x 3 x (LP3:LP0 + 1))	
1/4	Clock source/(1 x 4 x (LP3:LP0 + 1))	

TABLE 15-5: APPROXIMATE FRAME FREQUENCY (IN Hz) FOR LP PRESCALER SETTINGS

LP3:LP0	Multiplex Mode			
	Static	1/2	1/3	1/4
1	125	125	167	125
2	83	83	111	83
3	62	62	83	62
4	50	50	67	50
5	42	42	56	42
6	36	36	48	36
7	31	31	42	31

15.8 LCD Waveform Generation

LCD waveform generation is based on the principle that the net AC voltage across the dark pixel should be maximized and the net AC voltage across the clear pixel should be minimized. The net DC voltage across any pixel should be zero.

The COM signal represents the time slice for each common, while the SEG contains the pixel data. The pixel signal (COM-SEG) will have no DC component and it can take only one of the two rms values. The higher rms value will create a dark pixel and a lower rms value will create a clear pixel.

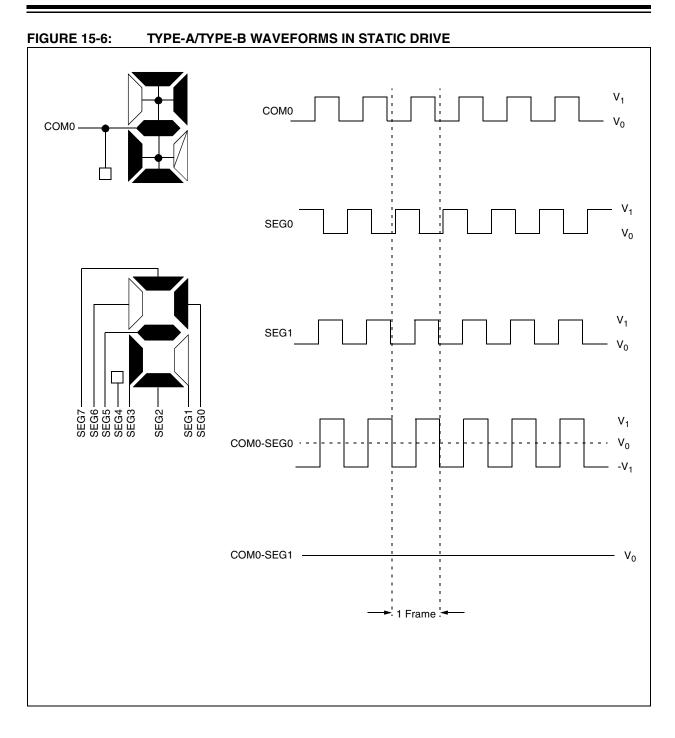
As the number of commons increases, the delta between the two rms values decreases. The delta represents the maximum contrast that the display can have.

The LCDs can be driven by two types of waveform: Type-A and Type-B. In the Type-A waveform, the phase changes within each common type, whereas in the Type-B waveform, the phase changes on each frame boundary. Thus, the Type-A waveform maintains 0 VDc over a single frame, whereas the Type-B waveform takes two frames.

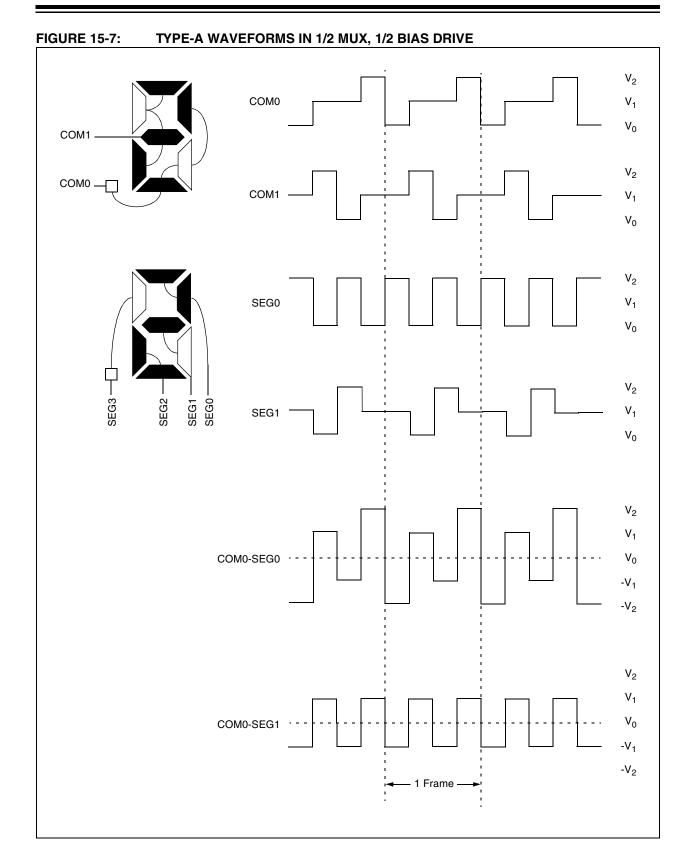
Note 1:	If the power-managed Sleep mode is
	invoked while the LCD Sleep bit is set
	(LCDCON<6> is '1'), take care to execute
	Sleep only when the VDC on all the pixels
	is '0'.

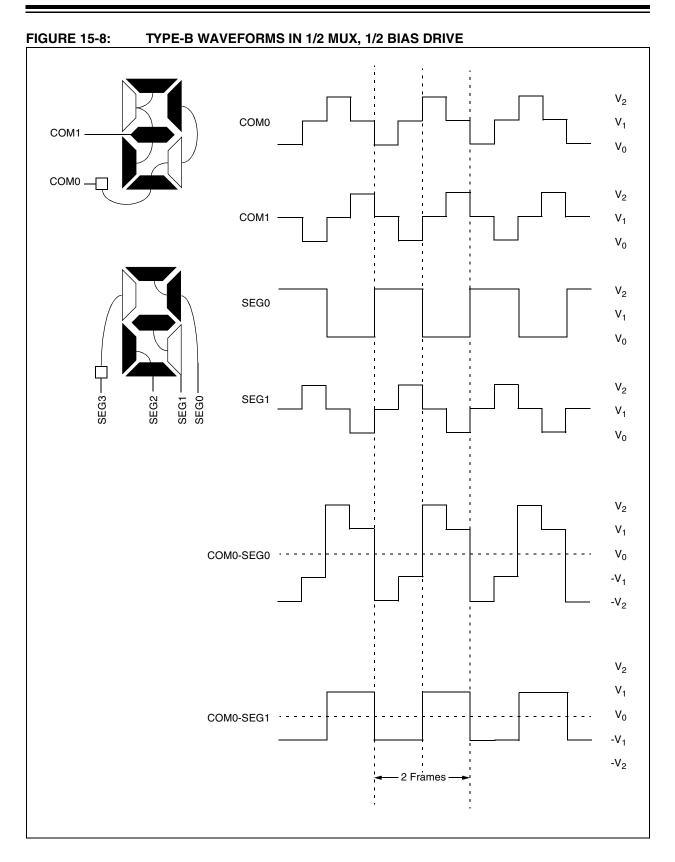
2: When the LCD clock source is the system clock, the LCD module will go to Sleep if the microcontroller goes into Sleep mode, regardless of the setting of the SPLEN bit. Thus, always take care to see that the VDC on all pixels is '0' whenever Sleep mode is invoked.

Figure 15-6 through Figure 15-16 provide waveforms for static, half multiplex, one-third multiplex and quarter multiplex drives for Type-A and Type-B waveforms.

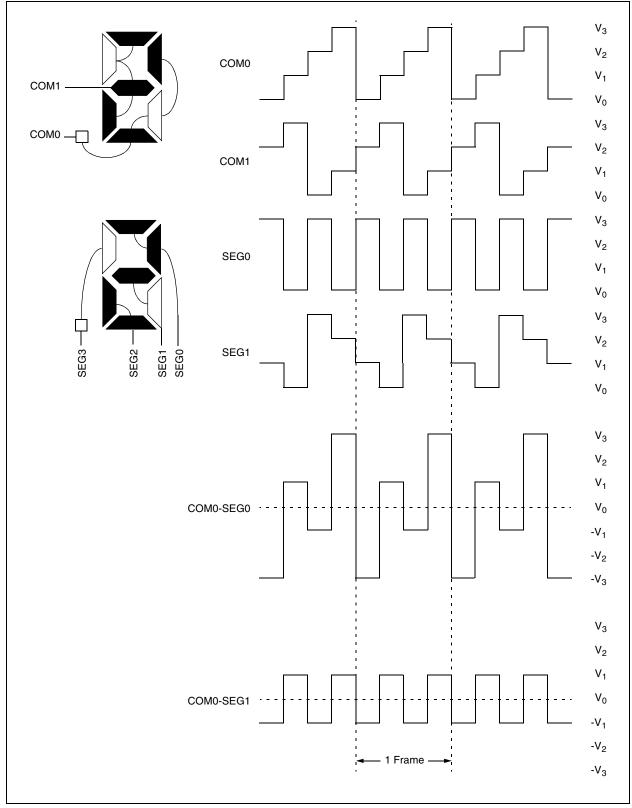


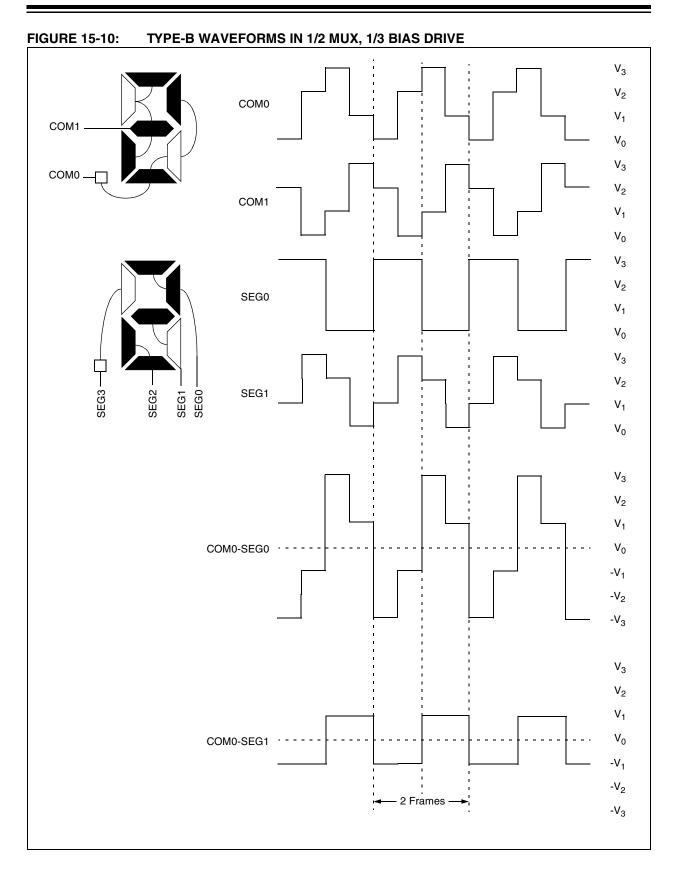
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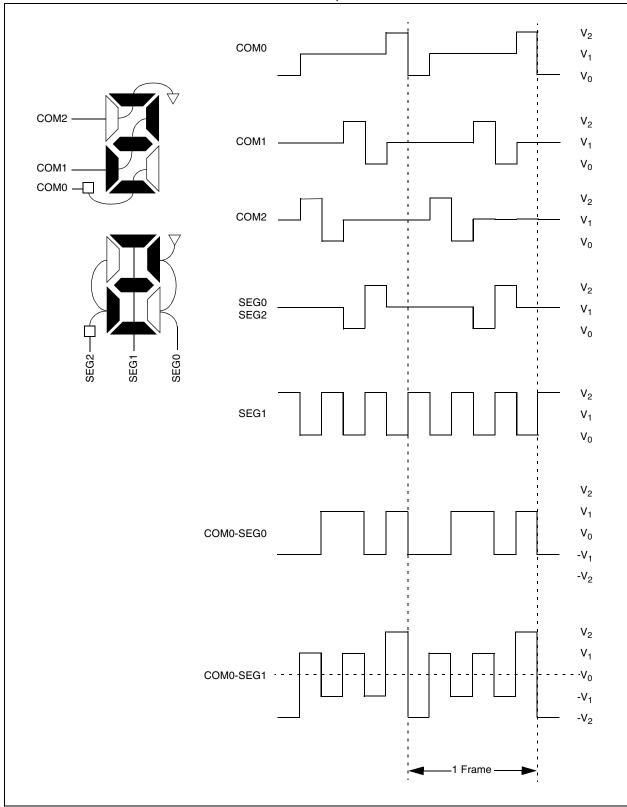
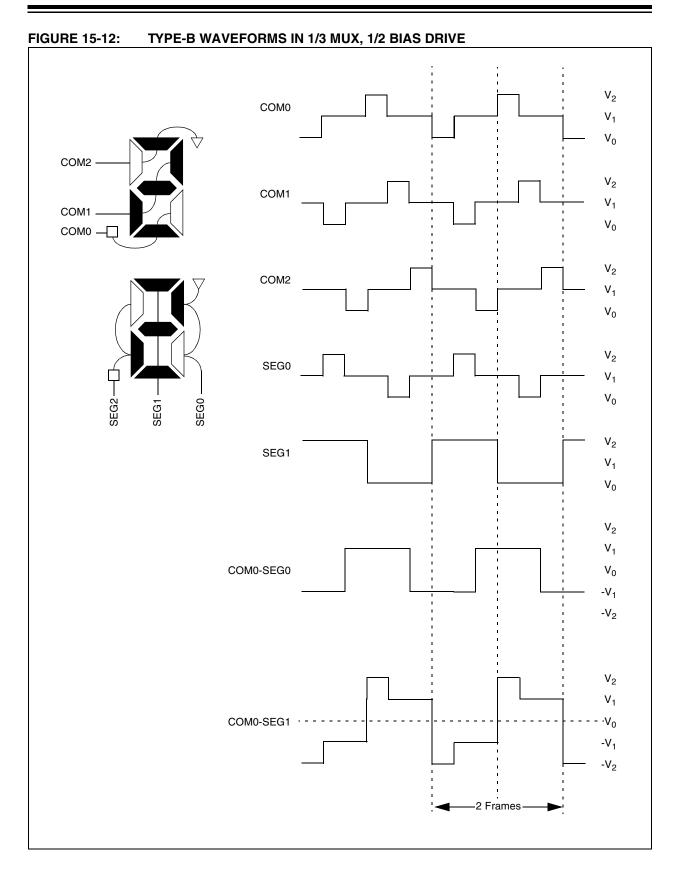
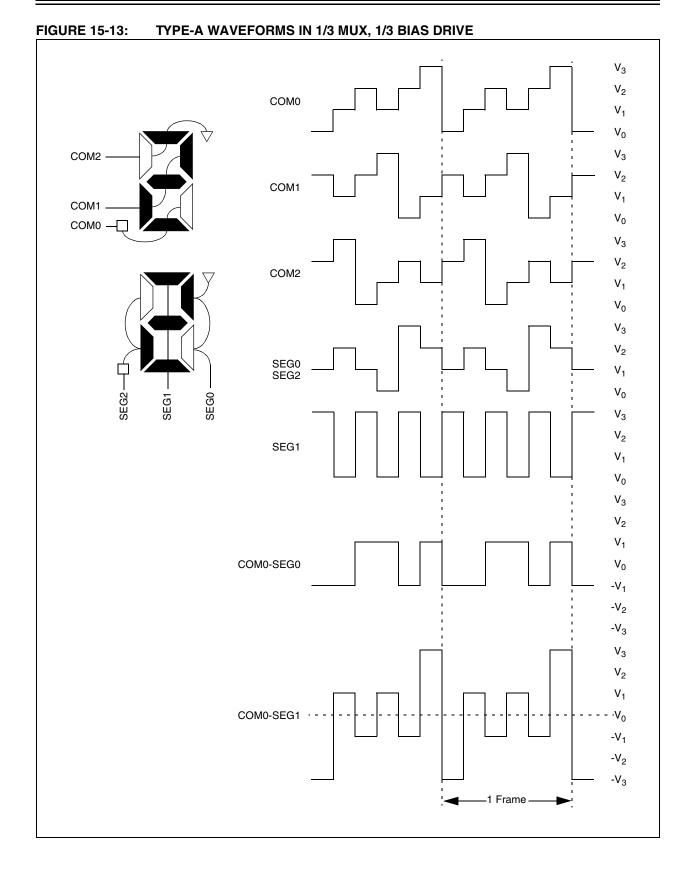
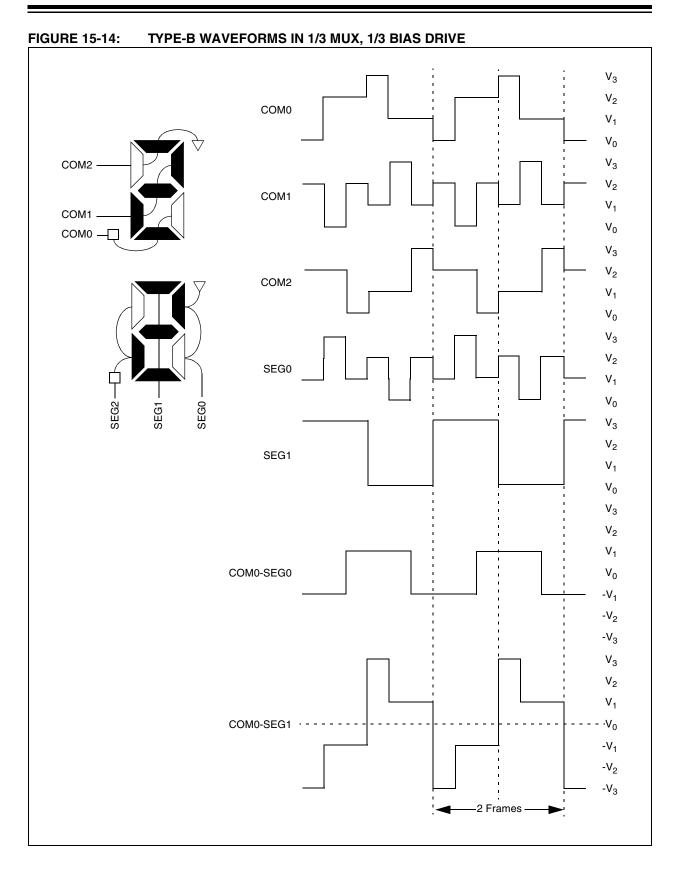


FIGURE 15-11: TYPE-A WAVEFORMS IN 1/3 MUX, 1/2 BIAS DRIVE



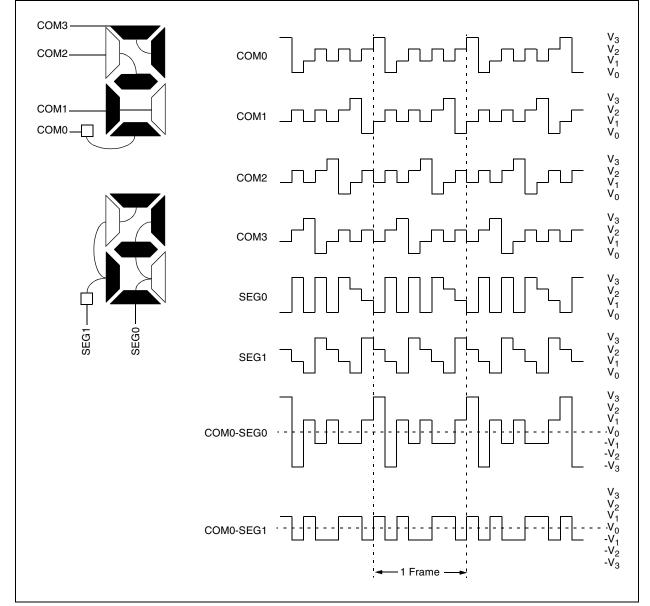


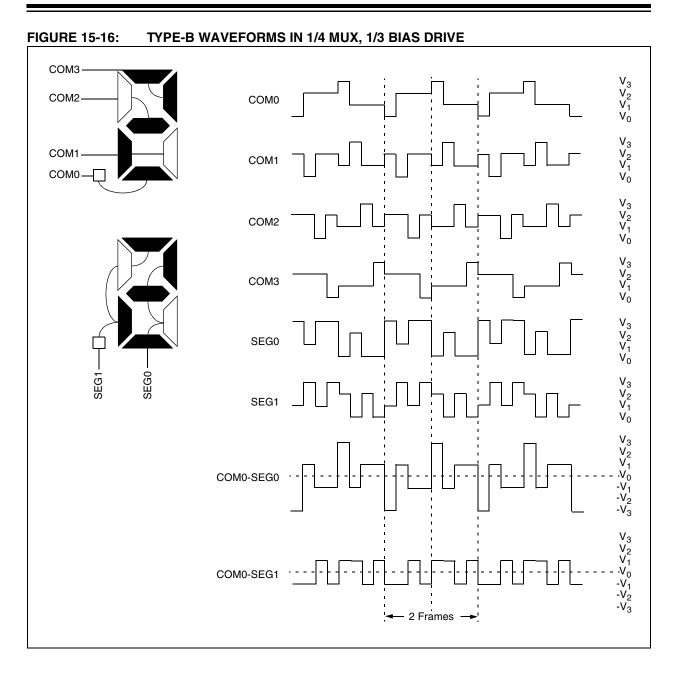
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15.9 LCD Interrupts

The LCD timing generation provides an interrupt that defines the LCD frame timing. This interrupt can be used to coordinate the writing of the pixel data with the start of a new frame. Writing pixel data at the frame boundary allows a visually crisp transition of the image. This interrupt can also be used to synchronize external events to the LCD. For example, the interface to an external segment driver can be synchronized for segment data update to the LCD frame.

A new frame is defined to begin at the leading edge of the COM0 common signal. The interrupt will be set immediately after the LCD controller completes accessing all pixel data required for a frame. This will occur at a fixed interval before the frame boundary (TFINT), as shown in Figure 15-17. The LCD controller will begin to access data for the next frame within the interval from the interrupt to when the controller begins to access data after the interrupt (TFWR). New data must be written within TFWR, as this is when the LCD controller will begin to access the data for the next frame. When the LCD driver is running with Type-B waveforms, and the LMUX1:LMUX0 bits are not equal to '00', there are some additional issues that must be addressed. Since the DC voltage on the pixel takes two frames to maintain zero volts, the pixel data must not change between subsequent frames. If the pixel data were allowed to change, the waveform for the odd frames would not necessarily be the complement of the waveform generated in the even frames and a DC component would be introduced into the panel. Therefore, when using Type-B waveforms, the user must synchronize the LCD pixel updates to occur within a subframe after the frame interrupt.

To correctly sequence writing while in Type-B, the interrupt will only occur on complete phase intervals. If the user attempts to write when the write is disabled, the WERR (LCDCON<5>) bit is set.

Note: The interrupt is not generated when the Type-A waveform is selected and when the Type-B with no multiplex (static) is selected.

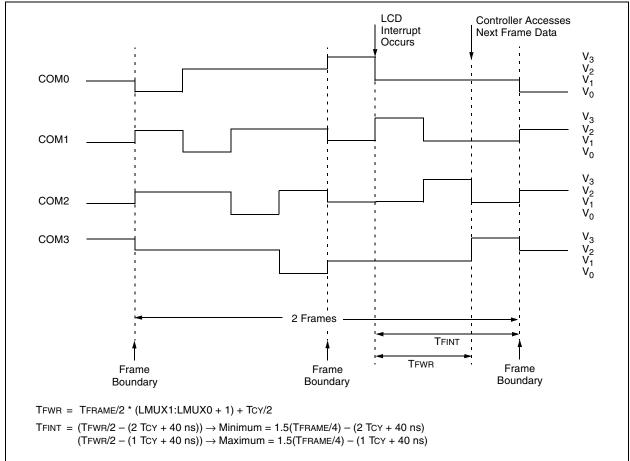


FIGURE 15-17: EXAMPLE WAVEFORMS AND INTERRUPT TIMING IN QUARTER DUTY CYCLE DRIVE

15.10 Operation During Sleep

The LCD module can operate during Sleep. The selection is controlled by the SLPEN bit (LCDCON<6>). Setting the SLPEN bit allows the LCD module to go to Sleep. Clearing the SLPEN bit allows the module to continue to operate during Sleep.

If a SLEEP instruction is executed and SLPEN = 1, the LCD module will cease all functions and go into a very low-current consumption mode. The module will stop operation immediately and drive the minimum LCD voltage on both segment and common lines. Figure 15-18 shows this operation.

To ensure that no DC component is introduced on the panel, the SLEEP instruction should be executed immediately after a LCD frame boundary. The LCD interrupt can be used to determine the frame boundary. See **Section 15.9 "LCD Interrupts**" for the formulas to calculate the delay.

If a SLEEP instruction is executed and SLPEN = 0, the module will continue to display the current contents of the LCDDATA registers. To allow the module to continue operation while in Sleep, the clock source must be either the Timer1 oscillator or one of the

internal oscillators (either INTRC or INTOSC as the default system clock). While in Sleep, the LCD data cannot be changed. The LCD module current consumption will not decrease in this mode; however, the overall consumption of the device will be lower due to shut down of the core and other peripheral functions.

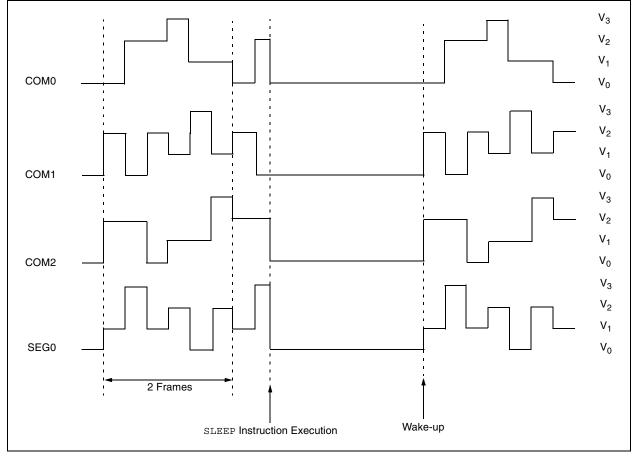
If the system clock is selected and the module is not configured for Sleep operation, the module will ignore the SLPEN bit and stop operation immediately. The minimum LCD voltage will then be driven onto the segments and commons

15.10.1 USING THE LCD REGULATOR DURING SLEEP

Applications that use the LCD regulator for bias generation may not achieve the same degree of power reductions in Sleep mode when compared to applications using Mode 3 (resistor ladder) biasing. This is particularly true with Mode 0 operation, where the charge pump is active.

If Modes 0, 1 or 2 are used for bias generation, software contrast control will not be available.





15.11 Configuring the LCD Module

The following is the sequence of steps to configure the LCD module.

- 1. Select the frame clock prescale using bits LP3:LP0 (LCDPS<3:0>).
- 2. Configure the appropriate pins to function as segment drivers using the LCDSEx registers.
- 3. Configure the appropriate pins as inputs using TRISx registers.
- 4. Configure the LCD module for the following using the LCDCON register:
 - Multiplex and Bias mode (LMUX1:LMUX0)
 - Timing source (CS1:CS0)
 - Sleep mode (SLPEN)
- 5. Write initial values to pixel data registers, LCDDATA0 through LCDDATA23.
- 6. Configure the LCD Regulator:
 - a) If M2 or M3 bias configuration is to be used, turn off the regulator by setting CKSEL<1:0> (LCDREG<1:0>) to '00'. Set or clear the CPEN bit (LCDREG<6>) to select Mode 2 or Mode 3, as appropriate.
 - b) If M0 or M1 bias generation is to be used:
 - Set the VBIAS level using the BIAS<2:0> bits (LCDREG<5:3>).
 - Set or clear the CPEN bit to enable or disable the charge pump.
 - Set or clear the MODE13 bit (LCDREG<2>) to select the Bias mode.
 - Select a regulator clock source using the CKSEL<1:0> bits.
- 7. Clear LCD Interrupt Flag, LCDIF (PIR3<6>), and if desired, enable the interrupt by setting the LCDIE bit (PIE3<6>).
- Enable the LCD module by setting the LCDEN bit (LCDCON<7>).

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51
PIR3		LCDIF	RC2IF	TX2IF		CCP2IF	CCP1IF	—	54
PIE3	_	LCDIE	RC2IE	TX2IE	_	CCP2IE	CCP1IE	_	54
IPR3	_	LCDIP	RC2IP	TX2IP		CCP2IP	CCP1IP	_	54
RCON	IPEN	—	_	RI	TO	PD	POR	BOR	52
LCDDATA23(1)	S47C3	S46C3	S45C3	S44C3	S43C3	S42C3	S41C3	S40C3	55
LCDDATA22	S39C3 ⁽¹⁾	S38C3 ⁽¹⁾	S37C3 ⁽¹⁾	S36C3 ⁽¹⁾	S35C3 ⁽¹⁾	S34C3 ⁽¹⁾	S33C3 ⁽¹⁾	S32C3	55
LCDDATA21	S31C3	S30C3	S29C3	S28C3	S27C3	S26C3	S25C3	S24C3	55
LCDDATA20	S23C3	S22C3	S21C3	S20C3	S19C3	S18C3	S17C3	S16C3	55
LCDDATA19	S15C3	S14C3	S13C3	S12C3	S11C3	S10C3	S09C3	S08C3	55
LCDDATA18	S07C3	S06C3	S05C3	S04C3	S03C3	S02C3	S01C3	S00C3	55
LCDDATA17(1)	S47C2	S46C2	S45C2	S44C2	S43C2	S42C2	S41C2	S40C2	55
LCDDATA16	S39C2 ⁽¹⁾	S38C2 ⁽¹⁾	S37C2 ⁽¹⁾	S36C2 ⁽¹⁾	S35C2 ⁽¹⁾	S34C2 ⁽¹⁾	S33C2 ⁽¹⁾	S32C2	55
LCDDATA15	S31C2	S30C2	S29C2	S28C2	S27C2	S26C2	S25C2	S24C2	55
LCDDATA14	S23C2	S22C2	S21C2	S20C2	S19C2	S18C2	S17C2	S16C2	55
LCDDATA13	S15C2	S14C2	S13C2	S12C2	S11C2	S10C2	S09C2	S08C2	55
LCDDATA12	S07C2	S06C2	S05C2	S04C2	S03C2	S02C2	S01C2	S00C2	55
LCDDATA11 ⁽¹⁾	S47C1	S46C1	S45C1	S44C1	S43C1	S42C1	S41C1	S40C1	55
LCDDATA10	S39C1 ⁽¹⁾	S38C1 ⁽¹⁾	S37C1 ⁽¹⁾	S36C1 ⁽¹⁾	S35C1 ⁽¹⁾	S34C1 ⁽¹⁾	S33C1 ⁽¹⁾	S32C1	55
LCDDATA9	S31C1	S30C1	S29C1	S28C1	S27C1	S26C1	S25C1	S24C1	55
LCDDATA8	S23C1	S22C1	S21C1	S20C1	S19C1	S18C1	S17C1	S16C1	55
LCDDATA7	S15C1	S14C1	S13C1	S12C1	S11C1	S10C1	S09C1	S08C1	55
LCDDATA6	S07C1	S06C1	S05C1	S04C1	S03C1	S02C1	S01C1	S00C1	55
LCDDATA5 ⁽¹⁾	S47C0	S46C0	S45C0	S44C0	S43C0	S42C0	S41C0	S40C0	55
LCDDATA4	S39C0 ⁽¹⁾	S38C0 ⁽¹⁾	S37C0 ⁽¹⁾	S36C0 ⁽¹⁾	S35C0 ⁽¹⁾	S34C0 ⁽¹⁾	S33C0 ⁽¹⁾	S32C0	53
LCDDATA3	S31C0	S30C0	S29C0	S28C0	S27C0	S26C0	S25C0	S24C0	53
LCDDATA2	S23C0	S22C0	S21C0	S20C0	S19C0	S18C0	S17C0	S16C0	53
LCDDATA1	S15C0	S14C0	S13C0	S12C0	S11C0	S10C0	S09C0	S08C0	53
LCDDATA0	S07C0	S06C0	S05C0	S04C0	S03C0	S02C0	S01C0	S00C0	53
LCDSE5 ⁽¹⁾	SE47	SE46	SE45	SE44	SE43	SE42	SE41	SE40	53
LCDSE4	SE39 ⁽¹⁾	SE38 ⁽¹⁾	SE37 ⁽¹⁾	SE36 ⁽¹⁾	SE35 ⁽¹⁾	SE34 ⁽¹⁾	SE33 ⁽¹⁾	SE32	53
LCDSE3	SE31	SE30	SE29	SE28	SE27	SE26	SE25	SE24	53
LCDSE2	SE23	SE22	SE21	SE20	SE19	SE18	SE17	SE16	53
LCDSE1	SE15	SE14	SE13	SE12	SE11	SE10	SE09	SE08	53
LCDSE0	SE07	SE06	SE05	SE04	SE03	SE02	SE01	SE00	53
LCDCON	LCDEN	SLPEN	WERR	—	CS1	CS0	LMUX1	LMUX0	53
LCDPS	WFT	BIASMD	LCDA	WA	LP3	LP2	LP1	LP0	53
LCDREG		CPEN	BIAS2	BIAS1	BIAS0	MODE13	CKSEL1	CKSEL0	

TABLE 15-6:	REGISTERS ASSOCIATED WITH LCD OPERATION
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Legend: — = unimplemented, read as '0'. Shaded cells are not used for LCD operation.

Note 1: These registers or individual bits are unimplemented on 64-pin devices.

NOTES:

16.0 MASTER SYNCHRONOUS SERIAL PORT (MSSP) MODULE

16.1 Master SSP (MSSP) Module Overview

The Master Synchronous Serial Port (MSSP) module is a serial interface, useful for communicating with other peripheral or microcontroller devices. These peripheral devices may be serial EEPROMs, shift registers, display drivers, A/D converters, etc. The MSSP module can operate in one of two modes:

- Serial Peripheral Interface (SPI)
- Inter-Integrated Circuit (I²C[™])
 - Full Master mode
 - Slave mode (with general address call)

The I^2C interface supports the following modes in hardware:

- Master mode
- Multi-Master mode
- Slave mode

16.2 Control Registers

Each MSSP module has three associated control registers. These include a status register (SSPSTAT) and two control registers (SSPCON1 and SSPCON2). The use of these registers and their individual bits differ significantly depending on whether the MSSP module is operated in SPI or I^2C mode.

Additional details are provided under the individual sections.

16.3 SPI Mode

The SPI mode allows 8 bits of data to be synchronously transmitted and received simultaneously. All four modes of SPI are supported. To accomplish communication, typically three pins are used:

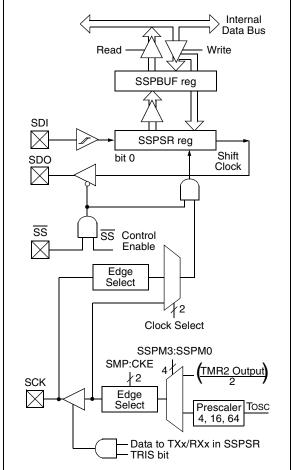
- Serial Data Out (SDO) RC5/SDO
- Serial Data In (SDI) RC4/SDI/SDA
- Serial Clock (SCK) RC3/SCK/SCL

Additionally, a fourth pin may be used when in a Slave mode of operation:

Slave Select (SS) – RF7/SS

Figure 16-1 shows the block diagram of the MSSP module when operating in SPI mode.





16.3.1 REGISTERS

Each MSSP module has four registers for SPI mode operation. These are:

- MSSP Control Register 1 (SSPCON1)
- MSSP Status Register (SSPSTAT)
- Serial Receive/Transmit Buffer Register (SSPBUF)
- MSSP Shift Register (SSPSR) Not directly accessible

SSPCON1 and SSPSTAT are the control and status registers in SPI mode operation. The SSPCON1 register is readable and writable. The lower 6 bits of the SSPSTAT are read-only. The upper two bits of the SSPSTAT are read/write. SSPSR is the shift register used for shifting data in or out. SSPBUF is the buffer register to which data bytes are written to or read from.

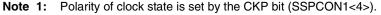
In receive operations, SSPSR and SSPBUF together create a double-buffered receiver. When SSPSR receives a complete byte, it is transferred to SSPBUF and the SSPIF interrupt is set.

During transmission, the SSPBUF is not double-buffered. A write to SSPBUF will write to both SSPBUF and SSPSR.

REGISTER 16-1: SSPSTAT: MSSP STATUS REGISTER (SPI MODE)

R/W-0	R/W-0	R-0	R-0	R-0	R-0	R0	R-0
SMP	CKE ⁽¹⁾	D/A	Р	S	R/W	UA	BF
bit 7							bit 0

Legend:				
R = Reada	able bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value	at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown
bit 7	SMP: San	nple bit		
	SPI Maste	r mode:		
	•	data sampled at end of data data sampled at middle of da	•	
	<u>SPI Slave</u> SMP musi	mode: t be cleared when SPI is use	ed in Slave mode.	
oit 6	CKE: SPI	Clock Select bit ⁽¹⁾		
		nit occurs on transition from nit occurs on transition from		
oit 5	D/A: Data	Address bit		
	Used in I ²	C™ mode only.		
oit 4	P: Stop bit	t		
	Used in I ²	C mode only. This bit is clea	ared when the MSSP module is	s disabled, SSPEN is cleared
oit 3	S: Start bi	t		
	Used in I ²	C mode only.		
bit 2	R/W: Rea	d/Write Information bit		
	Used in I ²	C mode only.		
bit 1	UA: Upda	te Address bit		
	Used in I ²	C mode only.		
bit 0	BF: Buffer	Full Status bit (Receive mo	de only)	
		ve complete, SSPBUF is ful		
	0 = Receiv	ve not complete, SSPBUF is	sempty	



R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
WCOL	SSPOV ⁽¹⁾	SSPEN ⁽²⁾	CKP	SSPM3 ⁽³⁾	SSPM2 ⁽³⁾	SSPM1 ⁽³⁾	SSPM0 ⁽³⁾
bit 7							bit 0
Legend:							
R = Readable		W = Writable b	oit		nented bit, read		
-n = Value at I	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkn	iown
bit 7	WCOL: Write	Collision Detec	t bit (Transmi	t mode onlv)			
		BUF register is	·	it is still transmi	tting the previo	ous word (must	t be cleared in
bit 6		eive Overflow In	dicator hit(1)				
	flow, the	te is received wh data in SSPSR even if only tra	is lost. Overf	UF register is st low can only oc a, to avoid settin	cur in Slave m	ode. The user	must read the
bit 5		ter Synchronous					
				K, SDO, SDI and se pins as I/O p		port pins	
bit 4	CKP: Clock F	Polarity Select bi	t				
		for clock is a hig for clock is a lov					
bit 3-0	SSPM3:SSPI	M0: Master Syn	chronous Ser	ial Port Mode S	elect bits ⁽³⁾		
	0100 = SPI S	lave mode, cloc	k = SCK pin,	SS pin control di SS pin control e		h be used as I/C) pin
	0010 = SPI M 0001 = SPI M	laster mode, clo laster mode, clo laster mode, clo laster mode, clo	ock = Fosc/64 ock = Fosc/16	1			

- **2:** When enabled, these pins must be properly configured as input or output.
- 3: Bit combinations not specifically listed here are either reserved or implemented in I²C[™] mode only.

16.3.2 OPERATION

When initializing the SPI, several options need to be specified. This is done by programming the appropriate control bits (SSPCON1<5:0> and SSPSTAT<7:6>). These control bits allow the following to be specified:

- Master mode (SCK is the clock output)
- Slave mode (SCK is the clock input)
- Clock Polarity (Idle state of SCK)
- Data Input Sample Phase (middle or end of data output time)
- Clock Edge (output data on rising/falling edge of SCK)
- Clock Rate (Master mode only)
- Slave Select mode (Slave mode only)

Each MSSP consists of a transmit/receive shift register (SSPSR) and a buffer register (SSPBUF). The SSPSR shifts the data in and out of the device, MSb first. The SSPBUF holds the data that was written to the SSPSR until the received data is ready. Once the 8 bits of data have been received, that byte is moved to the SSPBUF register. Then, the Buffer Full detect bit, BF (SSPSTAT<0>), and the interrupt flag bit, SSPIF, are set. This double-buffering of the received data (SSPBUF) allows the next byte to start reception before

reading the data that was just received. Any write to the SSPBUF register during transmission/reception of data will be ignored and the Write Collision detect bit, WCOL (SSPCON1<7>), will be set. User software must clear the WCOL bit so that it can be determined if the following write(s) to the SSPBUF register completed successfully.

When the application software is expecting to receive valid data, the SSPBUF should be read before the next byte of data to transfer is written to the SSPBUF. The Buffer Full bit, BF (SSPSTAT<0>), indicates when SSPBUF has been loaded with the received data (transmission is complete). When the SSPBUF is read, the BF bit is cleared. This data may be irrelevant if the SPI is only a transmitter. Generally, the MSSP interrupt is used to determine when the transmission/reception has completed. The SSPBUF must be read and/or written. If the interrupt method is not going to be used, then software polling can be done to ensure that a write collision does not occur. Example 16-1 shows the loading of the SSPBUF (SSPSR) for data transmission.

The SSPSR is not directly readable or writable and can only be accessed by addressing the SSPBUF register. Additionally, the SSPSTAT register indicates the various status conditions.

EXAMPLE 16-1: LOADING THE SSPBUF (SSPSR) REGISTER

LOOP	BTFSS	SSPSTAT, BF	;Has data been received (transmit complete)?
	BRA	LOOP	;No
	MOVF	SSPBUF, W	;WREG reg = contents of SSPBUF
	MOVWF	RXDATA	;Save in user RAM, if data is meaningful
	MOVF	TXDATA, W	;W reg = contents of TXDATA
	MOVWF	SSPBUF	;New data to xmit

16.3.3 ENABLING SPI I/O

To enable the serial port, MSSP Enable bit, SSPEN (SSPCON1<5>), must be set. To reset or reconfigure SPI mode, clear the SSPEN bit, reinitialize the SSPCON registers and then set the SSPEN bit. This configures the SDI, SDO, SCK and SS pins as serial port pins. For the pins to behave as the serial port function, some must have their data direction bits (in the TRIS register) appropriately programmed as follows:

- SDI is automatically controlled by the SPI module
- SDO must have TRISC<5> bit cleared
- SCK (Master mode) must have TRISC<3> bit cleared
- SCK (Slave mode) must have TRISC<3> bit set
- SS must have TRISF<7> bit set

Any serial port function that is not desired may be overridden by programming the corresponding data direction (TRIS) register to the opposite value.

16.3.4 OPEN-DRAIN OUTPUT OPTION

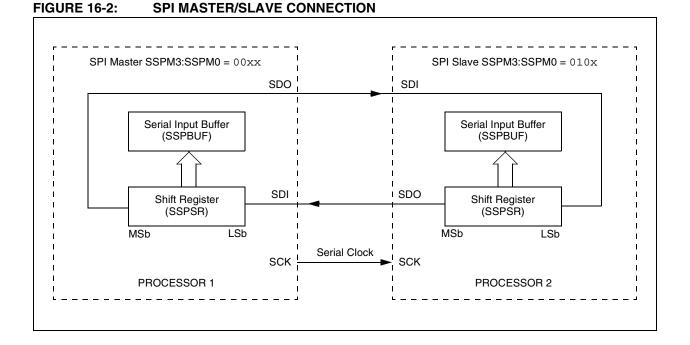
The drivers for the SDO output and SCK clock pins can be optionally configured as open-drain outputs. This feature allows the voltage level on the pin to be pulled to a higher level through an external pull-up resistor, and allows the output to communicate with external circuits without the need for additional level shifters.

The open-drain output option is controlled by the SPIOD bit (TRISG<7>). Setting the bit configures both pins for open-drain operation.

16.3.5 TYPICAL CONNECTION

Figure 16-2 shows a typical connection between two microcontrollers. The master controller (Processor 1) initiates the data transfer by sending the SCK signal. Data is shifted out of both shift registers on their programmed clock edge and latched on the opposite edge of the clock. Both processors should be programmed to the same Clock Polarity (CKP), then both controllers would send and receive data at the same time. Whether the data is meaningful (or dummy data) depends on the application software. This leads to three scenarios for data transmission:

- Master sends data Slave sends dummy data
- Master sends data Slave sends data
- Master sends dummy data Slave sends data



16.3.6 MASTER MODE

The master can initiate the data transfer at any time because it controls the SCK. The master determines when the slave (Processor 2, Figure 16-2) will broadcast data by the software protocol.

In Master mode, the data is transmitted/received as soon as the SSPBUF register is written to. If the SPI is only going to receive, the SDO output could be disabled (programmed as an input). The SSPSR register will continue to shift in the signal present on the SDI pin at the programmed clock rate. As each byte is received, it will be loaded into the SSPBUF register as if a normal received byte (interrupts and status bits appropriately set). This could be useful in receiver applications as a "Line Activity Monitor" mode. The clock polarity is selected by appropriately programming the CKP bit (SSPCON1<4>). This, then, would give waveforms for SPI communication as shown in Figure 16-3, Figure 16-5 and Figure 16-6, where the MSB is transmitted first. In Master mode, the SPI clock rate (bit rate) is user-programmable to be one of the following:

- Fosc/4 (or Tcy)
- Fosc/16 (or 4 Tcy)
- Fosc/64 (or 16 Tcy)
- Timer2 output/2

This allows a maximum data rate (at 40 MHz) of 10.00 Mbps.

Figure 16-3 shows the waveforms for Master mode. When the CKE bit is set, the SDO data is valid before there is a clock edge on SCK. The change of the input sample is shown based on the state of the SMP bit. The time when the SSPBUF is loaded with the received data is shown.

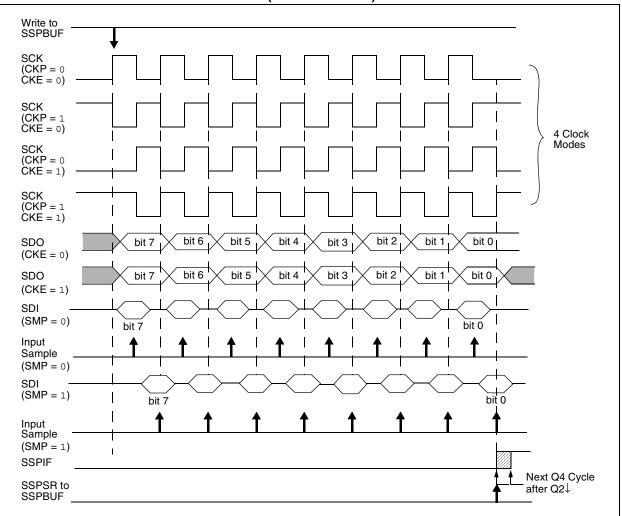


FIGURE 16-3: SPI MODE WAVEFORM (MASTER MODE)

16.3.7 SLAVE MODE

In Slave mode, the data is transmitted and received as the external clock pulses appear on SCK. When the last bit is latched, the SSPIF interrupt flag bit is set.

Before enabling the module in SPI Slave mode, the clock line must match the proper Idle state. The clock line can be observed by reading the SCK pin. The Idle state is determined by the CKP bit (SSPCON1<4>).

While in Slave mode, the external clock is supplied by the external clock source on the SCK pin. This external clock must meet the minimum high and low times as specified in the electrical specifications.

While in Sleep mode, the slave can transmit/receive data. When a byte is received, the device will wake-up from Sleep.

16.3.8 SLAVE SELECT SYNCHRONIZATION

The \overline{SS} pin allows a Synchronous Slave mode. The SPI must be in Slave mode with \overline{SS} pin control enabled (SSPCON1<3:0> = 04h). When the \overline{SS} pin is low, transmission and reception are enabled and the SDO pin is

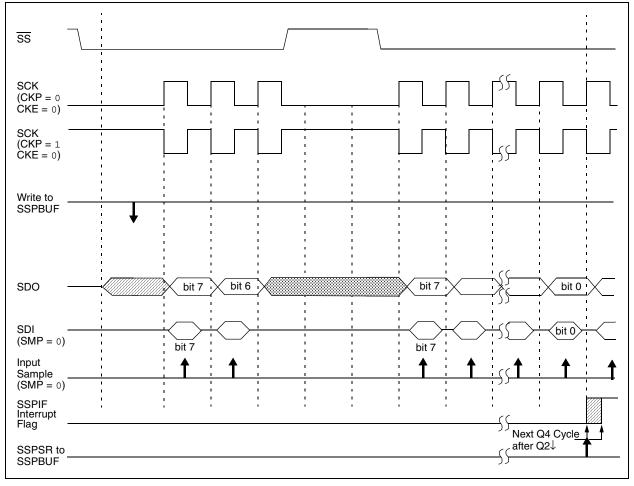
driven. When the \overline{SS} pin goes high, the SDO pin is no longer driven, even if in the middle of a transmitted byte and becomes a floating output. External pull-up/pull-down resistors may be desirable depending on the application.

- Note 1: When the SPI is in Slave mode with \overline{SS} pin control enabled (SSPCON1<3:0> = 0100), the SPI module will reset if the \overline{SS} pin is set to VDD.
 - 2: If the SPI is used in Slave mode with CKE set, then the SS pin control must be enabled.

When the SPI module resets, the bit counter is forced to '0'. This can be done by either forcing the \overline{SS} pin to a high level or clearing the SSPEN bit.

To emulate two-wire communication, the SDO pin can be connected to the SDI pin. When the SPI needs to operate as a receiver, the SDO pin can be configured as an input. This disables transmissions from the SDO. The SDI can always be left as an input (SDI function) since it cannot create a bus conflict.

FIGURE 16-4: SLAVE SYNCHRONIZATION WAVEFORM



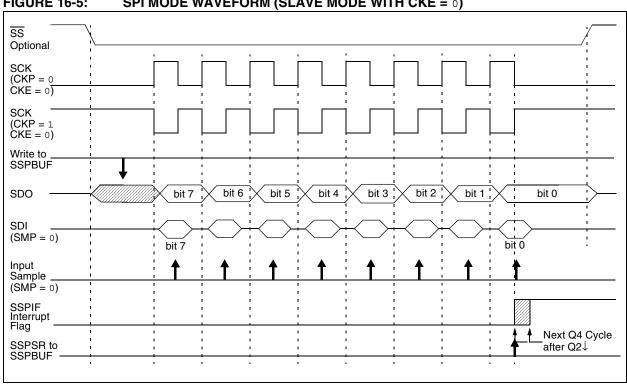
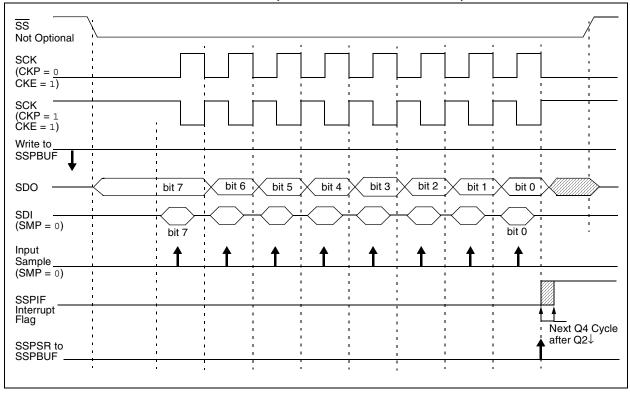


FIGURE 16-6: SPI MODE WAVEFORM (SLAVE MODE WITH CKE = 1)



16.3.9 OPERATION IN POWER-MANAGED MODES

In SPI Master mode, module clocks may be operating at a different speed than when in Full Power mode; in the case of Sleep mode, all clocks are halted.

In Idle modes, a clock is provided to the peripherals. That clock should be from the primary clock source, the secondary clock (Timer1 oscillator at 32.768 kHz) or the INTRC source. See **Section 2.3 "Clock Sources and Oscillator Switching"** for additional information.

In most cases, the speed that the master clocks SPI data is not important; however, this should be evaluated for each system.

If MSSP interrupts are enabled, they can wake the controller from Sleep mode, or one of the Idle modes, when the master completes sending data. If an exit from Sleep or Idle mode is not desired, MSSP interrupts should be disabled.

If the Sleep mode is selected, all module clocks are halted and the transmission/reception will remain in that state until the devices wakes. After the device returns to Run mode, the module will resume transmitting and receiving data.

In SPI Slave mode, the SPI Transmit/Receive Shift register operates asynchronously to the device. This allows the device to be placed in any power-managed mode and data to be shifted into the SPI Transmit/Receive Shift register. When all 8 bits have been received, the MSSP interrupt flag bit will be set and if enabled, will wake the device.

16.3.10 EFFECTS OF A RESET

A Reset disables the MSSP module and terminates the current transfer.

16.3.11 BUS MODE COMPATIBILITY

Table 16-1 shows the compatibility between the standard SPI modes and the states of the CKP and CKE control bits.

TABLE 16-1:	SPI BUS MODES

Standard SPI Mode	Control E	Bits State
Terminology	СКР	CKE
0, 0	0	1
0, 1	0	0
1, 0	1	1
1, 1	1	0

There is also an SMP bit which controls when the data is sampled.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INTOIE	RBIE	TMR0IF	INT0IF	RBIF	51
PIR1	—	ADIF	RC1IF	TX1IF	SSPIF	—	TMR2IF	TMR1IF	54
PIE1	—	ADIE	RC1IE	TX1IE	SSPIE	—	TMR2IE	TMR1IE	54
IPR1	—	ADIP	RC1IP	TX1IP	SSPIP	—	TMR2IP	TMR1IP	54
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	54
TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1		54
TRISG	SPIOD	CCP2OD	CCP10D	TRISG4	TRISG3	TRISG2	TRISG1	TRISG0	54
SSPBUF	MSSP Rec	eive Buffer/1	Fransmit Re	gister					52
SSPCON1	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	52
SSPSTAT	SMP	CKE	D/A	Р	S	R/W	UA	BF	52

TABLE 16-2: REGISTERS ASSOCIATED WITH SPI OPERATION

Legend: Shaded cells are not used by the MSSP module in SPI mode.

16.4 I²C Mode

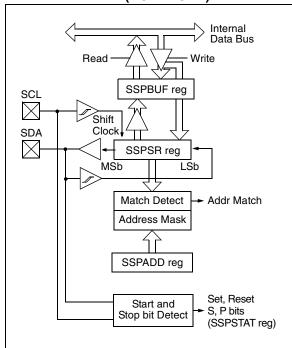
The MSSP module in I²C mode fully implements all master and slave functions (including general call support) and provides interrupts on Start and Stop bits in hardware to determine a free bus (multi-master function). The MSSP module implements the standard mode specifications as well as 7-bit and 10-bit addressing.

Two pins are used for data transfer:

- Serial clock (SCL) RC3/SCK/SCL
- Serial data (SDA) RC4/SDI/SDA

The user must configure these pins as inputs by setting the TRISC<4:3> bits.

FIGURE 16-7: MSSP BLOCK DIAGRAM (I²C™ MODE)



16.4.1 REGISTERS

The MSSP module has six registers for $\mathsf{I}^2\mathsf{C}$ operation. These are:

- MSSP Control Register 1 (SSPCON1)
- MSSP Control Register 2 (SSPCON2)
- MSSP Status Register (SSPSTAT)
- Serial Receive/Transmit Buffer Register (SSPBUF)
- MSSP Shift Register (SSPSR) Not directly accessible
- MSSP Address Register (SSPADD)

SSPCON1, SSPCON2 and SSPSTAT are the control and status registers in l^2C mode operation. The SSPCON1 and SSPCON2 registers are readable and writable. The lower 6 bits of the SSPSTAT are read-only. The upper two bits of the SSPSTAT are read/write.

Many of the bits in SSPCON2 assume different functions, depending on whether the module is operating in Master or Slave mode; bits <5:2> also assume different names in Slave mode. The different aspects of SSPCON2 are shown in Register 16-5 (for Master mode) and Register 16-6 (Slave mode).

SSPSR is the shift register used for shifting data in or out. SSPBUF is the buffer register to which data bytes are written to or read from.

SSPADD register holds the slave device address when the MSSP is configured in I^2C Slave mode. When the MSSP is configured in Master mode, the lower seven bits of SSPADD act as the Baud Rate Generator reload value.

In receive operations, SSPSR and SSPBUF together create a double-buffered receiver. When SSPSR receives a complete byte, it is transferred to SSPBUF and the SSPIF interrupt is set.

During transmission, the SSPBUF is not double-buffered. A write to SSPBUF will write to both SSPBUF and SSPSR.

R/W-0	R/W-0	R-0	R-0	R-0	R-0	R0	R-0
SMP	CKE	D/A	P ⁽¹⁾	S ⁽¹⁾	R/W	UA	BF
bit 7		•		•			bit (
Legend:							
R = Read	able bit	W = Writable	bit	U = Unimplem	ented bit, rea	ıd as '0'	
-n = Value	e at POR	'1' = Bit is set		'0' = Bit is clea		x = Bit is unkno	wn
bit 7	SMP: Slew F	ate Control bit					
	In Master or						
				d Speed mode (1 MHz)	
L H 0			ed for High-Sp	eed mode (400	KHZ)		
bit 6	CKE: SMBus In Master or						
		MBus specific i	nputs				
		SMBus specific	nputs				
bit 5	D/A: Data/Ad	ldress bit					
	<u>In Master mo</u> Reserved.	de:					
	In Slave mod	le:					
				transmitted was			
		that the last by	te received or	transmitted was	address		
bit 4	P: Stop bit ⁽¹⁾						
		that a Stop bit vas not detected		cted last			
bit 3	S: Start bit ⁽¹⁾						
		that a Start bit vas not detected		cted last			
bit 2	R/W: Read/V	Vrite Informatior	ı bit (I ² C mode	e only)			
	In Slave mod		,				
	1 = Read						
	0 = Write	(3)					
	<u>In Master mo</u> 1 = Transmit	is in progress					
		is not in progre	SS				
bit 1	UA: Update	Address bit (10-	Bit Slave mod	e only)			
				e the address in	the SSPADD	register	
		does not need t	o be updated				
bit 0	BF: Buffer Fi						
	<u>In Transmit n</u> 1 = SSPBUF						
	0 = SSPBUF						
	In Receive m	ode:					
				CK and Stop bit			
	0 = SSPBUF	is empty (does	not include th	e ACK and Stop	DITS)		
Note 1:	This bit is cleared	d on Reset and	when SSPEN	is cleared.			
2:	This bit holds the address match to				s match. This	bit is only valid fr	om the
3:	ORing this bit wit		=		dicate if the N	ISSP is in Active	mode.

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0			
WCOL	SSPOV	SSPEN ⁽¹⁾	CKP	SSPM3	SSPM2	SSPM1	SSPM0			
bit 7	·						bit (
Legend:										
R = Readab		W = Writable		-	nented bit, read					
-n = Value a	It POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown			
bit 7	WCOL: Write	Collision Dete	ct bit							
	In Master Tra									
	1 = A write t	to the SSPBUR				nditions were i	not valid for			
		sion to be starte	ed (must be cle	eared in softwa	re)					
	0 = No collis									
	In Slave Tran 1 = The SSF software	BUF register is	s written while	it is still transm	itting the previo	ous word (mus	t be cleared i			
	0 = No collis									
	<u>In Receive m</u> This is a "don	<u>ode (Master or</u> i't care" bit.	Slave modes)	<u>.</u>						
bit 6	SSPOV: Receive Overflow Indicator bit									
	<u>In Receive m</u> 1 = A byte is software 0 = No overf	received while)	the SSPBUF	register is still h	olding the prev	ious byte (mus	t be cleared i			
	In Transmit m		ansmit mode.							
bit 5	SSPEN: Mas	ter Synchronou	is Serial Port E	Enable bit ⁽¹⁾						
	1 = Enables t	he serial port a serial port a	nd configures	the SDA and S		serial port pins				
bit 4	CKP: SCK R	elease Control	bit							
	In Slave mod									
	1 = Release 0	clock ck low (clock st	rotob) used to	oncura data c	atun timo					
	In Master mo	•	reich), useu io	ensure data se	etup time					
	Unused in thi									
bit 3-0	SSPM3:SSP	M0: Synchrono	us Serial Port	Mode Select bi	ts					
	$1111 = I^{2}C S$ $1110 = I^{2}C S$ $1011 = I^{2}C F$ $1000 = I^{2}C N$	lave mode, 10- lave mode, 7-b irmware Contro laster mode, clo lave mode, 10-	bit address wit it address with olled Master mo ock = Fosc/(4	h Start and Sto Start and Stop ode (slave Idle)	p bit interrupts bit interrupts e					
		lave mode, 70-								
	Bit combinatio									

REGISTER 16-4: SSPCON1: MSSP CONTROL REGISTER 1 (I²C[™] MODE)

Note 1: When enabled, the SDA and SCL pins must be configured as inputs.

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
GCEN	ACKSTAT	ACKDT ⁽¹⁾	ACKEN ⁽²⁾	RCEN ⁽²⁾	PEN ⁽²⁾	RSEN ⁽²⁾	SEN ⁽²⁾
bit 7							bit
Legend:							
R = Readab	ble bit	W = Writable	bit	U = Unimplem	nented bit, rea	d as '0'	
-n = Value a	at POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkn	own
bit 7		eral Call Enable	bit				
	Unused in Ma						
bit 6		•		Transmit mode	e only)		
		edge was not re edge was receiv		ave			
bit 5		•		ceive mode only	_{v)} (1)		
bit 0	1 = Not Ackn				y)		
	0 = Acknowle	0					
bit 4	ACKEN: Ack	nowledge Sequ	lence Enable b	it (2)			
	1 = Initiate A	.cknowledge se	quence on SD/	A and SCL pins	and transmit /	ACKDT data bit.	Automatical
		by hardware.					
		edge sequence					
bit 3		•		e mode only) ⁽²⁾			
	1 = Enables I 0 = Receive I	Receive mode t	or I ² C				
bit 2		ondition Enable	hit(2)				
	•			. pins. Automati	ically cleared I	by hardware.	
	0 = Stop cond			- pe. / teree.		,	
bit 1	RSEN: Repe	ated Start Cond	lition Enable bi	t ⁽²⁾			
	1 = Initiate R	Repeated Start of	condition on SE	A and SCL pin	s. Automatica	lly cleared by ha	ardware.
	•	d Start conditio					
bit 0		ondition Enable					
	1 = Initiate St 0 = Start cone		SDA and SCL	pins. Automat	ically cleared I	by hardware.	
Note 1: \	Value that will be	transmitted wh	en the user ini	tiates an Ackno	wledae seaue	ence at the end o	of a receive.
	120 11				• .		

2: If the I²C module is active, these bits may not be set (no spooling) and the SSPBUF may not be written (or writes to the SSPBUF are disabled).

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0					
GCEN	ACKSTAT	ADMSK5	ADMSK4	ADMSK3	ADMSK2	ADMSK1	SEN ⁽¹⁾					
bit 7							bit (
Legend:												
R = Readable bit		W = Writable bit $U = Unimplemented bit, read as '0'$										
-n = Value at POR		'1' = Bit is set		'0' = Bit is cleared		x = Bit is unknown						
bit 7	GCEN: General Call Enable bit											
	1 = Enable interrupt when a general call address (0000h) is received in the SSPSR											
	0 = General call address disabled											
bit 6	ACKSTAT: Acknowledge Status bit											
	Unused in Slave mode.											
bit 5-2	ADMSK5:ADMSK2: Slave Address Mask Select bits											
	1 = Masking of corresponding bits of SSPADD enabled											
	0 = Masking of corresponding bits of SSPADD disabled											
bit 1	ADMSK1: Slave Address Least Significant bit(s) Mask Select bit											
bit 1	In 7-Bit Address mode:											
bit 1						1 = Masking of SSPADD<1> only enabled 0 = Masking of SSPADD<1> only disabled						
bit 1	1 = Masking o	of SSPADD<1>										
bit 1	1 = Masking o 0 = Masking o	of SSPADD<1> of SSPADD<1>										
bit 1	1 = Masking c 0 = Masking c In 10-Bit Addr	of SSPADD<1> of SSPADD<1> ress mode:	only disabled									
bit 1	1 = Masking o 0 = Masking o <u>In 10-Bit Addr</u> 1 = Masking o	of SSPADD<1> of SSPADD<1> ress mode: of SSPADD<1:(only disabled									
	1 = Masking o 0 = Masking o <u>In 10-Bit Addr</u> 1 = Masking o 0 = Masking o	of SSPADD<1> of SSPADD<1> ress mode: of SSPADD<1:(of SSPADD<1:(only disabled									
bit 1 bit 0	1 = Masking o 0 = Masking o <u>In 10-Bit Addr</u> 1 = Masking o 0 = Masking o SEN: Stretch	of SSPADD<1> of SSPADD<1> <u>ress mode:</u> of SSPADD<1:(of SSPADD<1:(Enable bit ⁽¹⁾	only disabled)> enabled)> disabled	ve transmit and	l slave receive	(stretch enable	d)					

Note 1: If the I²C module is active, this bit may not be set (no spooling) and the SSPBUF may not be written (or writes to the SSPBUF are disabled).

16.4.2 OPERATION

The MSSP module functions are enabled by setting the MSSP Enable bit, SSPEN (SSPCON1<5>).

The SSPCON1 register allows control of the I^2C operation. Four mode selection bits (SSPCON1<3:0>) allow one of the following I^2C modes to be selected:

- I²C Master mode, clock = (FOSC/4) x (SSPADD + 1)
- I²C Slave mode (7-bit address)
- I²C Slave mode (10-bit address)
- I²C Slave mode (7-bit address) with Start and Stop bit interrupts enabled
- I²C Slave mode (10-bit address) with Start and Stop bit interrupts enabled
- I²C Firmware Controlled Master mode, slave is Idle

Selection of any I²C mode, with the SSPEN bit set, forces the SCL and SDA pins to be open-drain, provided these pins are programmed to inputs by setting the appropriate TRISC or TRISD bits. To ensure proper operation of the module, pull-up resistors must be provided externally to the SCL and SDA pins.

16.4.3 SLAVE MODE

In Slave mode, the SCL and SDA pins must be configured as inputs (TRISC<4:3> set). The MSSP module will override the input state with the output data when required (slave-transmitter).

The I²C Slave mode hardware will always generate an interrupt on an exact address match. In addition, address masking will also allow the hardware to generate an interrupt for more than one address (up to 31 in 7-bit addressing and up to 63 in 10-bit addressing). Through the mode select bits, the user can also choose to interrupt on Start and Stop bits.

When an address is matched, or the data transfer after an address match is received, the hardware automatically will generate the Acknowledge (\overline{ACK}) pulse and load the SSPBUF register with the received value currently in the SSPSR register.

Any combination of the following conditions will cause the MSSP module not to give this ACK pulse:

- The Buffer Full bit, BF (SSPSTAT<0>), was set before the transfer was received.
- The overflow bit, SSPOV (SSPCON1<6>), was set before the transfer was received.

In this case, the SSPSR register value is not loaded into the SSPBUF, but bit SSPIF is set. The BF bit is cleared by reading the SSPBUF register, while bit SSPOV is cleared through software. The SCL clock input must have a minimum high and low for proper operation. The high and low times of the I^2C specification, as well as the requirement of the MSSP module, are shown in timing parameter 100 and parameter 101.

16.4.3.1 Addressing

Once the MSSP module has been enabled, it waits for a Start condition to occur. Following the Start condition, the 8 bits are shifted into the SSPSR register. All incoming bits are sampled with the rising edge of the clock (SCL) line. The value of register SSPSR<7:1> is compared to the value of the SSPADD register. The address is compared on the falling edge of the eighth clock (SCL) pulse. If the addresses match and the BF and SSPOV bits are clear, the following events occur:

- 1. The SSPSR register value is loaded into the SSPBUF register.
- 2. The Buffer Full bit, BF, is set.
- 3. An ACK pulse is generated.
- 4. The MSSP Interrupt Flag bit, SSPIF, is set (and interrupt is generated, if enabled) on the falling edge of the ninth SCL pulse.

In 10-Bit Address mode, two address bytes need to be received by the slave. The five Most Significant bits (MSbs) of the first address byte specify if this is a 10-bit address. Bit R/W (SSPSTAT<2>) must specify a write so the slave device will receive the second address byte. For a 10-bit address, the first byte would equal '11110 A9 A8 0', where 'A9' and 'A8' are the two MSbs of the address. The sequence of events for 10-bit address is as follows, with steps 7 through 9 for the slave-transmitter:

- 1. Receive first (high) byte of address (bits SSPIF, BF and UA (SSPSTAT<1>) are set).
- Update the SSPADD register with second (low) byte of address (clears bit UA and releases the SCL line).
- 3. Read the SSPBUF register (clears bit BF) and clear flag bit, SSPIF.
- 4. Receive second (low) byte of address (SSPIF, BF and UA bits are set).
- 5. Update the SSPADD register with the first (high) byte of address. If match releases SCL line, this will clear UA bit.
- 6. Read the SSPBUF register (clears bit BF) and clear flag bit, SSPIF.
- 7. Receive Repeated Start condition.
- 8. Receive first (high) byte of address (SSPIF and BF bits are set).
- 9. Read the SSPBUF register (clears BF bit) and clear flag bit, SSPIF.

16.4.3.2 Address Masking

Masking an address bit causes that bit to become a "don't care". When one address bit is masked, two addresses will be Acknowledged and cause an interrupt. It is possible to mask more than one address bit at a time, which makes it possible to Acknowledge up to 31 addresses in 7-bit mode and up to 63 addresses in 10-bit mode (see Example 16-2).

The I^2C Slave behaves the same way whether address masking is used or not. However, when address masking is used, the I^2C slave can Acknowledge multiple addresses and cause interrupts. When this occurs, it is necessary to determine which address caused the interrupt by checking SSPBUF.

In 7-Bit Address mode, address mask bits, ADMSK<5:1> (SSPCON<5:1>), mask the corresponding address bits in the SSPADD register. For any ADMSK bits that are set (ADMSK<n> = 1), the corresponding address bit is ignored (SSPADD<n> = x). For the module to issue an address Acknowledge, it is sufficient to match only on addresses that do not have an active address mask.

EXAMPLE 16-2: ADDRESS MASKING EXAMPLES

In 10-Bit Address mode, ADMSK<5:2> bits mask the corresponding address bits in the SSPADD register. In addition, ADMSK1 simultaneously masks the two LSbs of the address (SSPADD<1:0>). For any ADMSK bits that are active (ADMSK<n> = 1), the corresponding address bit is ignored (SSPADD<n> = x). Also note that although in 10-Bit Addressing mode, the upper address bits reuse part of the SSPADD register bits, the address mask bits do not interact with those bits. They only affect the lower address bits.

Note 1: ADMSK1 masks the two Least Significant bits of the address.

2: The two Most Significant bits of the address are not affected by address masking.

7-Bit Addressing:

SSPADD<7:1> = A0h (1010000) (SSPADD<0> is assumed to be '0')

ADMSK<5:1> = 00111

Addresses Acknowledged: A0h, A2h, A4h, A6h, A8h, AAh, ACh, AEh

10-Bit Addressing:

SSPADD<7:0> = A0h (10100000) (the two MSbs of the address are ignored in this example, since they are not affected by masking)

ADMSK<5:1> = 00111

Addresses Acknowledged: A0h, A1h, A2h, A3h, A4h, A5h, A6h, A7h, A8h, A9h, AAh, ABh, ACh, ADh, AEh, AFh

16.4.3.3 Reception

When the R/\overline{W} bit of the address byte is clear and an address match occurs, the R/\overline{W} bit of the SSPSTAT register is cleared. The received address is loaded into the SSPBUF register and the SDA line is held low (ACK).

When the address byte overflow condition exists, then the no Acknowledge (ACK) pulse is given. An overflow condition is defined as either bit, BF (SSPSTAT<0>), is set or bit, SSPOV (SSPCON1<6>), is set.

An MSSP interrupt is generated for each data transfer byte. The interrupt flag bit, SSPIF, must be cleared in software. The SSPSTAT register is used to determine the status of the byte.

If SEN is enabled (SSPCON2<0> = 1), SCK/SCL will be held low (clock stretch) following each data transfer. The clock must be released by setting bit, CKP (SSPCON1<4>). See **Section 16.4.4** "Clock **Stretching**" for more details.

16.4.3.4 Transmission

When the R/\overline{W} bit of the incoming address byte is set and an address match occurs, the R/\overline{W} bit of the SSPSTAT register is set. The received address is loaded into the SSPBUF register. The ACK pulse will be sent on the ninth bit and pin RC3 is held low, regardless of SEN (see **Section 16.4.4 "Clock Stretching"** for more details). By stretching the clock, the master will be unable to assert another clock pulse until the slave is done preparing the transmit data. The transmit data must be loaded into the SSPBUF register which also loads the SSPSR register. Then, pin RC3 should be enabled by setting bit, CKP (SSPCON1<4>). The eight data bits are shifted out on the falling edge of the SCL input. This ensures that the SDA signal is valid during the SCL high time (Figure 16-10).

The ACK pulse from the master-receiver is latched on the rising edge of the ninth SCL input pulse. If the SDA line is high (not ACK), then the data transfer is complete. In this case, when the ACK is latched by the slave, the slave logic is reset (resets SSPSTAT register) and the slave monitors for another occurrence of the Start bit. If the SDA line was low (ACK), the next transmit data must be loaded into the SSPBUF register. Again, pin RC3 must be enabled by setting bit, CKP.

An MSSP interrupt is generated for each data transfer byte. The SSPIF bit must be cleared in software and the SSPSTAT register is used to determine the status of the byte. The SSPIF bit is set on the falling edge of the ninth clock pulse.

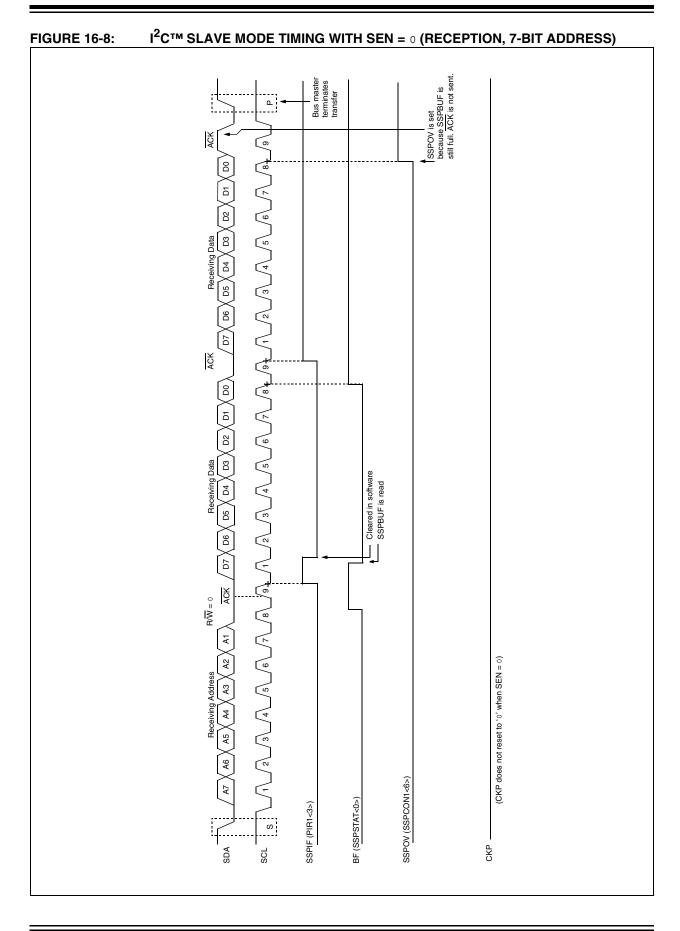
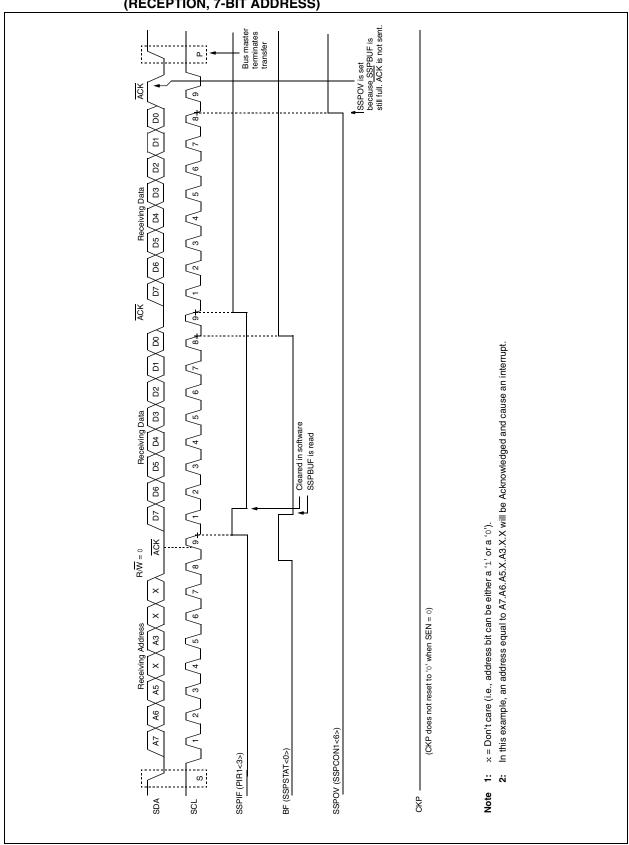
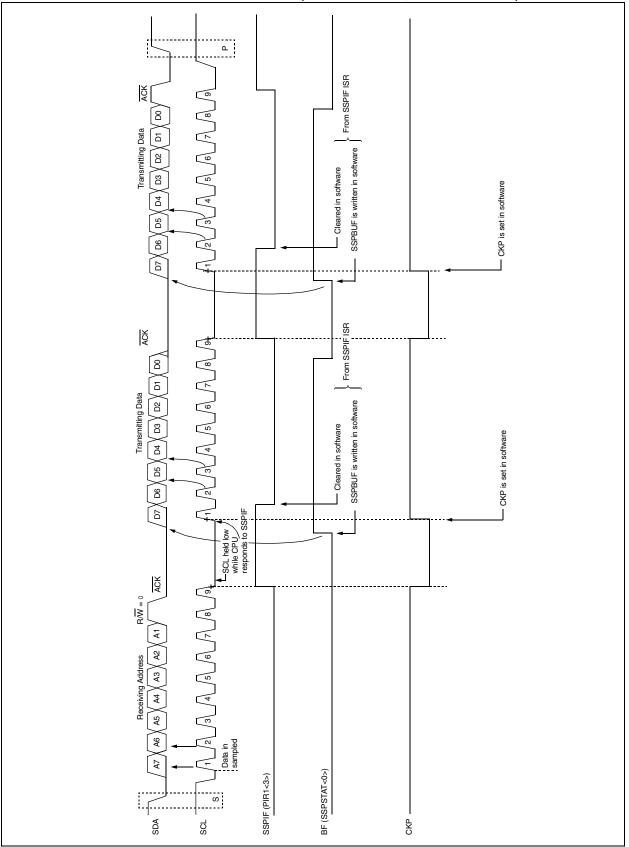


FIGURE 16-9: I²C[™] SLAVE MODE TIMING WITH SEN = 0 AND ADMSK<5:1> = 01011 (RECEPTION, 7-BIT ADDRESS)







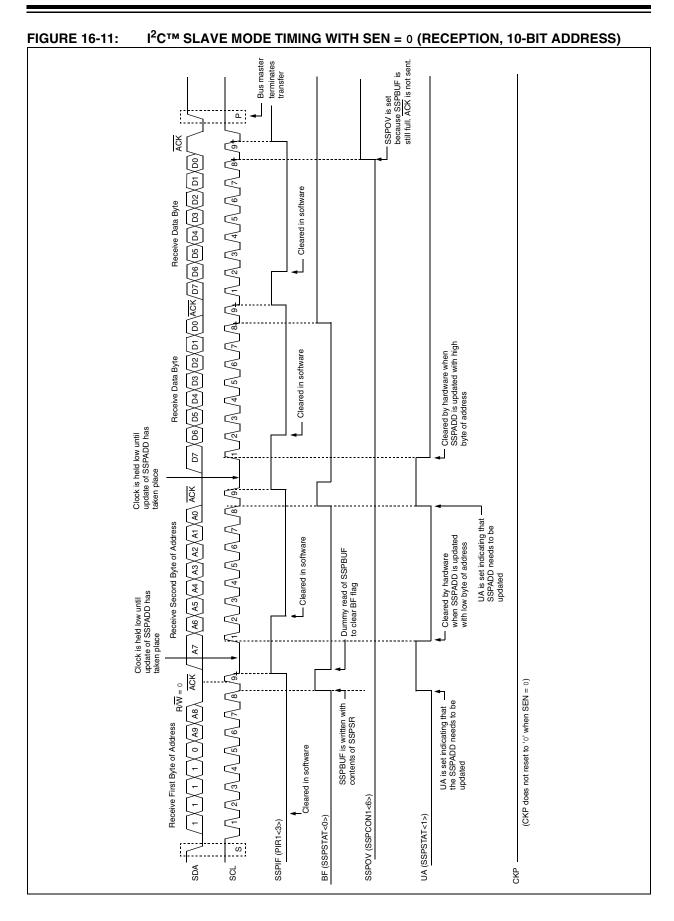
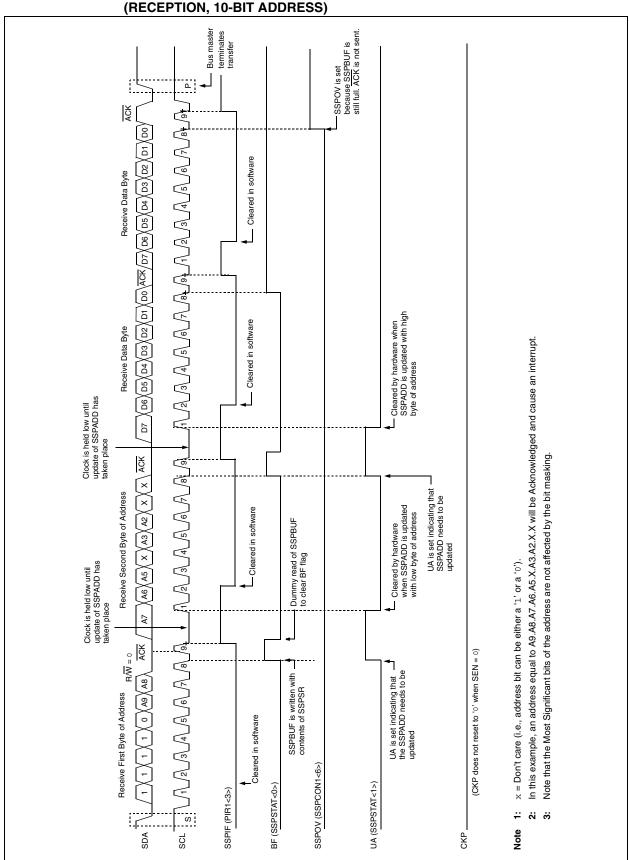
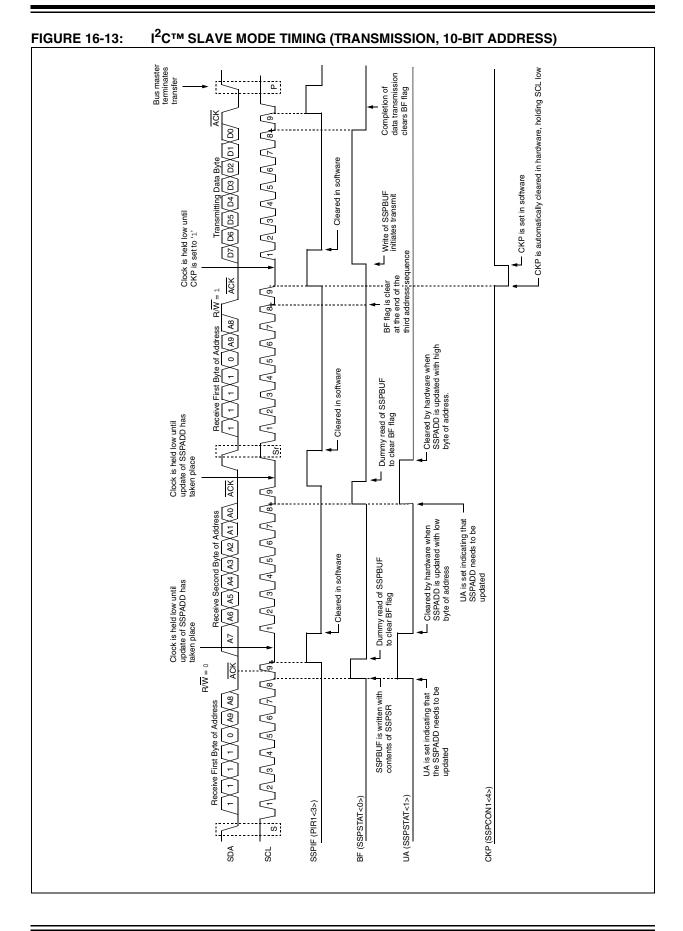


FIGURE 16-12:



I²C[™] SLAVE MODE TIMING WITH SEN = 0 AND ADMSK<5:1> = 01001

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16.4.4 CLOCK STRETCHING

Both 7-Bit and 10-Bit Slave modes implement automatic clock stretching during a transmit sequence.

The SEN bit (SSPCON2<0>) allows clock stretching to be enabled during receives. Setting SEN will cause the SCL pin to be held low at the end of each data receive sequence.

16.4.4.1 Clock Stretching for 7-Bit Slave Receive Mode (SEN = 1)

In 7-Bit Slave Receive mode, on the falling edge of the ninth clock at the end of the ACK sequence, if the BF bit is set, the CKP bit in the SSPCON1 register is automatically cleared, forcing the SCL output to be held low. The CKP being cleared to '0' will assert the SCL line low. The CKP bit must be set in the user's ISR before reception is allowed to continue. By holding the SCL line low, the user has time to service the ISR and read the contents of the SSPBUF before the master device can initiate another receive sequence. This will prevent buffer overruns from occurring (see Figure 16-15).

- Note 1: If the user reads the contents of the SSPBUF before the falling edge of the ninth clock, thus clearing the BF bit, the CKP bit will not be cleared and clock stretching will not occur.
 - 2: The CKP bit can be set in software regardless of the state of the BF bit. The user should be careful to clear the BF bit in the ISR before the next receive sequence in order to prevent an overflow condition.

16.4.4.2 Clock Stretching for 10-Bit Slave Receive Mode (SEN = 1)

In 10-Bit Slave Receive mode, during the address sequence, clock stretching automatically takes place but CKP is not cleared. During this time, if the UA bit is set after the ninth clock, clock stretching is initiated. The UA bit is set after receiving the upper byte of the 10-bit address and following the receive of the second byte of the 10-bit address with the R/W bit cleared to '0'. The release of the clock line occurs upon updating SSPADD. Clock stretching will occur on each data receive sequence as described in 7-bit mode.

Note: If the user polls the UA bit and clears it by updating the SSPADD register before the falling edge of the ninth clock occurs and if the user hasn't cleared the BF bit by reading the SSPBUF register before that time, then the CKP bit will still NOT be asserted low. Clock stretching on the basis of the state of the BF bit only occurs during a data sequence, not an address sequence.

16.4.4.3 Clock Stretching for 7-Bit Slave Transmit Mode

The 7-Bit Slave Transmit mode implements clock stretching by clearing the CKP bit after the falling edge of the ninth clock if the BF bit is clear. This occurs regardless of the state of the SEN bit.

The user's ISR must set the CKP bit before transmission is allowed to continue. By holding the SCL line low, the user has time to service the ISR and load the contents of the SSPBUF before the master device can initiate another transmit sequence (see Figure 16-10).

- Note 1: If the user loads the contents of SSPBUF, setting the BF bit before the falling edge of the ninth clock, the CKP bit will not be cleared and clock stretching will not occur.
 - 2: The CKP bit can be set in software regardless of the state of the BF bit.

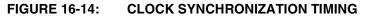
16.4.4.4 Clock Stretching for 10-Bit Slave Transmit Mode

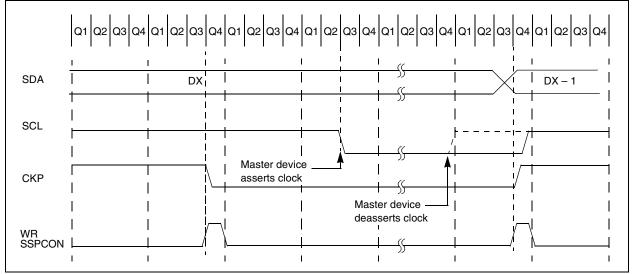
In 10-Bit Slave Transmit mode, clock stretching is controlled during the first two address sequences by the state of the UA bit, just as it is in 10-Bit Slave Receive mode. The first two addresses are followed by a third address sequence which contains the high-order bits of the 10-bit address and the R/W bit set to '1'. After the third address sequence is performed, the UA bit is not set, the module is now configured in Transmit mode and clock stretching is controlled by the BF flag as in 7-Bit Slave Transmit mode (see Figure 16-13).

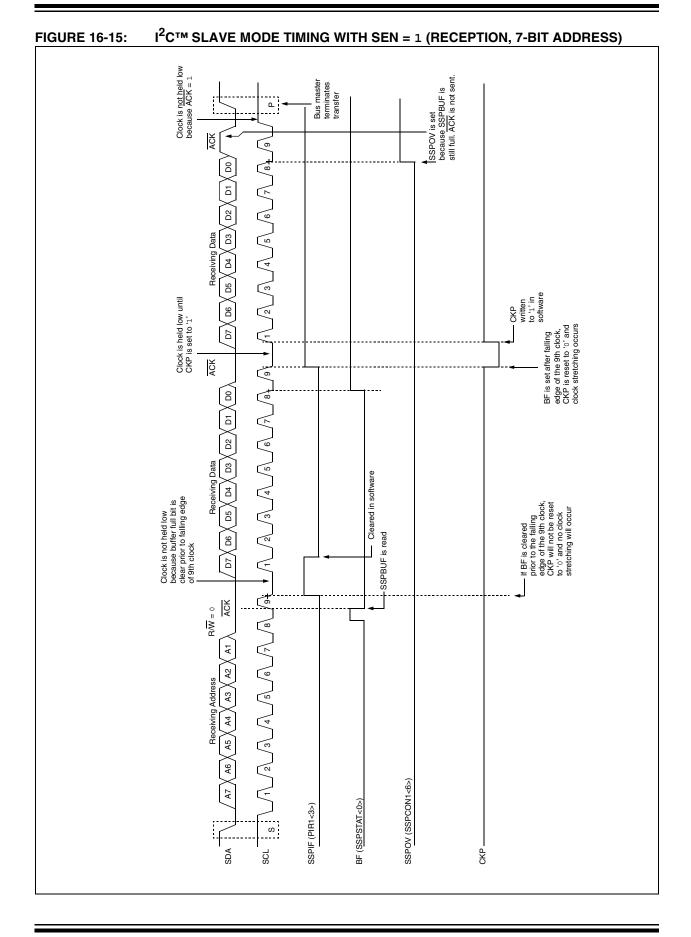
16.4.4.5 Clock Synchronization and the CKP bit

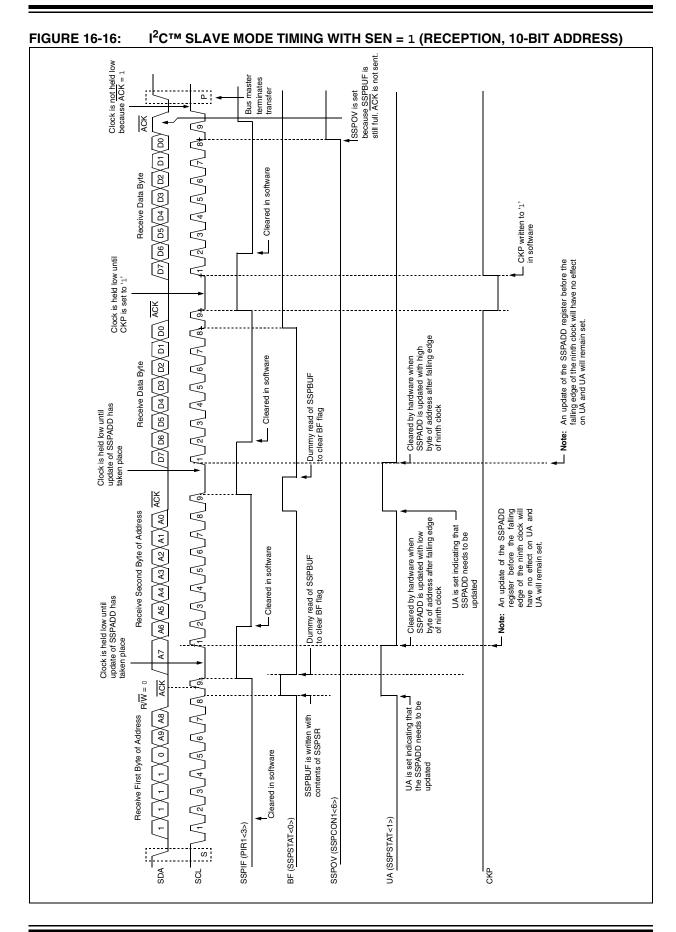
When the CKP bit is cleared, the SCL output is forced to '0'. However, clearing the CKP bit will not assert the SCL output low until the SCL output is already sampled low. Therefore, the CKP bit will not assert the SCL line until an external I^2C master device has

already asserted the SCL line. The SCL output will remain low until the CKP bit is set and all other devices on the I^2 C bus have deasserted SCL. This ensures that a write to the CKP bit will not violate the minimum high time requirement for SCL (see Figure 16-14).









16.4.5 GENERAL CALL ADDRESS SUPPORT

The addressing procedure for the I^2C bus is such that the first byte after the Start condition usually determines which device will be the slave addressed by the master. The exception is the general call address which can address all devices. When this address is used, all devices should, in theory, respond with an Acknowledge.

The general call address is one of eight addresses reserved for specific purposes by the I^2C protocol. It consists of all '0's with R/W = 0.

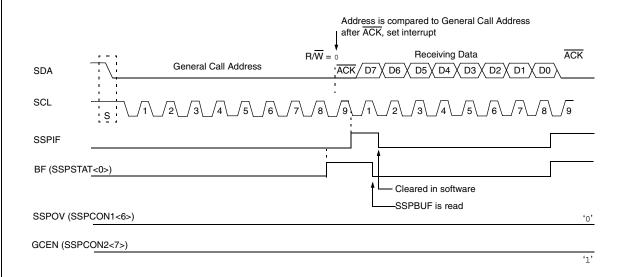
The general call address is recognized when the General Call Enable bit, GCEN, is enabled (SSPCON2<7> set). Following a Start bit detect, 8 bits are shifted into the SSPSR and the address is compared against the SSPADD. It is also compared to the general call address and fixed in hardware.

If the general call address matches, the SSPSR is transferred to the SSPBUF, the BF flag bit is set (eighth bit) and on the falling edge of the ninth bit (ACK bit), the SSPIF interrupt flag bit is set.

When the interrupt is serviced, the source for the interrupt can be checked by reading the contents of the SSPBUF. The value can be used to determine if the address was device specific or a general call address.

In 10-bit mode, the SSPADD is required to be updated for the second half of the address to match and the UA bit is set (SSPSTAT<1>). If the general call address is sampled when the GCEN bit is set, while the slave is configured in 10-Bit Address mode, then the second half of the address is not necessary, the UA bit will not be set and the slave will begin receiving data after the Acknowledge (Figure 16-17).





16.4.6 MASTER MODE

Master mode is enabled by setting and clearing the appropriate SSPM bits in SSPCON1 and by setting the SSPEN bit. In Master mode, the SCL and SDA lines are manipulated by the MSSP hardware.

Master mode of operation is supported by interrupt generation on the detection of the Start and Stop conditions. The Stop (P) and Start (S) bits are cleared from a Reset or when the MSSP module is disabled. Control of the I^2C bus may be taken when the P bit is set, or the bus is Idle, with both the S and P bits clear.

In Firmware Controlled Master mode, user code conducts all I^2C bus operations based on Start and Stop bit conditions.

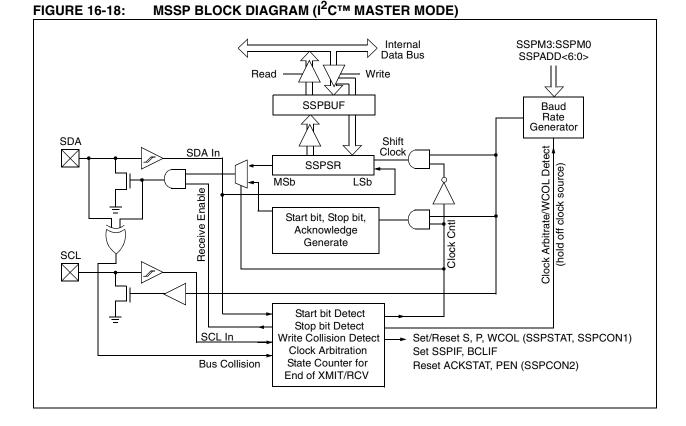
Once Master mode is enabled, the user has six options.

- 1. Assert a Start condition on SDA and SCL.
- 2. Assert a Repeated Start condition on SDA and SCL.
- 3. Write to the SSPBUF register initiating transmission of data/address.
- 4. Configure the I²C port to receive data.
- 5. Generate an Acknowledge condition at the end of a received byte of data.
- 6. Generate a Stop condition on SDA and SCL.

Note: The MSSP module, when configured in I²C Master mode, does not allow queueing of events. For instance, the user is not allowed to initiate a Start condition and immediately write the SSPBUF register to initiate transmission before the Start condition is complete. In this case, the SSPBUF will not be written to and the WCOL bit will be set, indicating that a write to the SSPBUF did not occur.

The following events will cause the MSSP Interrupt Flag bit, SSPIF, to be set (and MSSP interrupt, if enabled):

- Start condition
- Stop condition
- Data transfer byte transmitted/received
- Acknowledge transmit
- Repeated Start



16.4.6.1 I²C Master Mode Operation

The master device generates all of the serial clock pulses and the Start and Stop conditions. A transfer is ended with a Stop condition or with a Repeated Start condition. Since the Repeated Start condition is also the beginning of the next serial transfer, the I²C bus will not be released.

In Master Transmitter mode, serial data is output through SDA, while SCL outputs the serial clock. The first byte transmitted contains the slave address of the receiving device (7 bits) and the Read/Write (R/W) bit. In this case, the R/W bit will be logic '0'. Serial data is transmitted 8 bits at a time. After each byte is transmitted, an Acknowledge bit is received. Start and Stop conditions are output to indicate the beginning and the end of a serial transfer.

In Master Receive mode, the first byte transmitted contains the slave address of the transmitting device (7 bits) and the R/W bit. In this case, the R/W bit will be logic '1'. Thus, the first byte transmitted is a 7-bit slave address followed by a '1' to indicate the receive bit. Serial data is received via SDA, while SCL outputs the serial clock. Serial data is received 8 bits at a time. After each byte is received, an Acknowledge bit is transmitted. Start and Stop conditions indicate the beginning and end of transmission.

The Baud Rate Generator used for the SPI mode operation is used to set the SCL clock frequency for either 100 kHz, 400 kHz or 1 MHz I²C operation. See **Section 16.4.7 "Baud Rate"** for more detail.

A typical transmit sequence would go as follows:

- 1. The user generates a Start condition by setting the Start Enable bit, SEN (SSPCON2<0>).
- 2. SSPIF is set. The MSSP module will wait the required start time before any other operation takes place.
- 3. The user loads the SSPBUF with the slave address to transmit.
- 4. Address is shifted out the SDA pin until all 8 bits are transmitted.
- 5. The MSSP module shifts in the ACK bit from the slave device and writes its value into the SSPCON2 register (SSPCON2<6>).
- 6. The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSPIF bit.
- 7. The user loads the SSPBUF with eight bits of data.
- 8. Data is shifted out the SDA pin until all 8 bits are transmitted.
- The MSSP module shifts in the ACK bit from the slave device and writes its value into the SSPCON2 register (SSPCON2<6>).
- 10. The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSPIF bit.
- 11. The user generates a Stop condition by setting the Stop Enable bit, PEN (SSPCON2<2>).
- 12. Interrupt is generated once the Stop condition is complete.

16.4.7 BAUD RATE

In I²C Master mode, the Baud Rate Generator (BRG) reload value is placed in the lower 7 bits of the SSPADD register (Figure 16-19). When a write occurs to SSPBUF, the Baud Rate Generator will automatically begin counting. The BRG counts down to '0' and stops until another reload has taken place. The BRG count is decremented twice per instruction cycle (TcY) on the Q2 and Q4 clocks. In I²C Master mode, the BRG is reloaded automatically.

Once the given operation is complete (i.e., transmission of the last data bit is followed by \overline{ACK}), the internal clock will automatically stop counting and the SCL pin will remain in its last state.

Table 16-3 demonstrates clock rates based on instruction cycles and the BRG value loaded into SSPADD.

16.4.7.1 Baud Rate Generation in Power-Managed Modes

When the device is operating in one of the power-managed modes, the clock source to the BRG may change frequency or even stop, depending on the mode and clock source selected. Switching to a Run or Idle mode from either the secondary clock or internal oscillator is likely to change the clock rate to the BRG. In Sleep mode, the BRG will not be clocked at all.

FIGURE 16-19: BAUD RATE GENERATOR BLOCK DIAGRAM

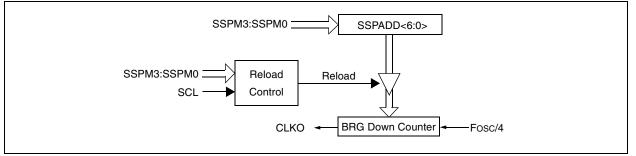


TABLE 16-3: I²C[™] CLOCK RATE w/BRG

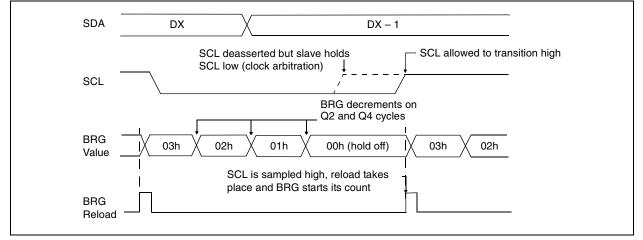
Fcy	Fcy * 2	BRG Value	FSCL (2 Rollovers of BRG)
10 MHz	20 MHz	18h	400 kHz ⁽¹⁾
10 MHz	20 MHz	1Fh	312.5 kHz
10 MHz	20 MHz	63h	100 kHz
4 MHz	8 MHz	09h	400 kHz ⁽¹⁾
4 MHz	8 MHz	0Ch	308 kHz
4 MHz	8 MHz	27h	100 kHz
1 MHz	2 MHz	02h	333 kHz ⁽¹⁾
1 MHz	2 MHz	09h	100 kHz
1 MHz	2 MHz	00h	1 MHz ⁽¹⁾

Note 1: The I²C[™] interface does not conform to the 400 kHz I²C specification (which applies to rates greater than 100 kHz) in all details, but may be used with care where higher rates are required by the application.

16.4.7.2 Clock Arbitration

Clock arbitration occurs when the master, during any receive, transmit or Repeated Start/Stop condition, deasserts the SCL pin (SCL allowed to float high). When the SCL pin is allowed to float high, the Baud Rate Generator (BRG) is suspended from counting until the SCL pin is actually sampled high. When the SCL pin is sampled high, the Baud Rate Generator is reloaded with the contents of SSPADD<6:0> and begins counting. This ensures that the SCL high time will always be at least one BRG rollover count in the event that the clock is held low by an external device (Figure 16-20).





16.4.8 I²C MASTER MODE START CONDITION TIMING

To initiate a Start condition, the user sets the Start Enable bit, SEN (SSPCON2<0>). If the SDA and SCL pins are sampled high, the Baud Rate Generator is reloaded with the contents of SSPADD<6:0> and starts its count. If SCL and SDA are both sampled high when the Baud Rate Generator times out (TBRG), the SDA pin is driven low. The action of the SDA being driven low while SCL is high is the Start condition and causes the S bit (SSPSTAT<3>) to be set. Following this, the Baud Rate Generator is reloaded with the contents of SSPADD<6:0> and resumes its count. When the Baud Rate Generator times out (TBRG), the SEN bit (SSPCON2<0>) will be automatically cleared by hardware. The Baud Rate Generator is suspended, leaving the SDA line held low and the Start condition is complete.

Note: If at the beginning of the Start condition, the SDA and SCL pins are already sampled low, or if during the Start condition, the SCL line is sampled low before the SDA line is driven low, a bus collision occurs. The Bus Collision Interrupt Flag, BCLIF, is set, the Start condition is aborted and the I²C module is reset into its Idle state.

16.4.8.1 WCOL Status Flag

If the user writes the SSPBUF when a Start sequence is in progress, the WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).

Note: Because queueing of events is not allowed, writing to the lower 5 bits of SSPCON2 is disabled until the Start condition is complete.

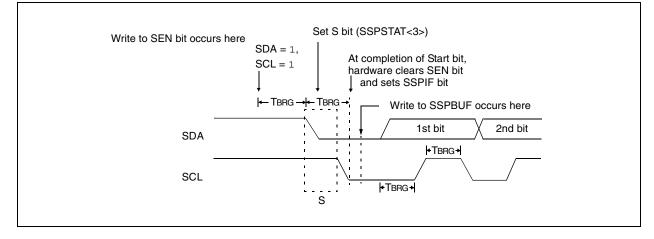


FIGURE 16-21: FIRST START BIT TIMING

16.4.9 I²C MASTER MODE REPEATED START CONDITION TIMING

A Repeated Start condition occurs when the RSEN bit (SSPCON2<1>) is programmed high and the I²C logic module is in the Idle state. When the RSEN bit is set, the SCL pin is asserted low. When the SCL pin is sampled low, the Baud Rate Generator is loaded with the contents of SSPADD<6:0> and begins counting. The SDA pin is released (brought high) for one Baud Rate Generator count (TBRG). When the Baud Rate Generator times out, if SDA is sampled high, the SCL pin will be deasserted (brought high). When SCL is sampled high, the Baud Rate Generator is reloaded with the contents of SSPADD<6:0> and begins counting. SDA and SCL must be sampled high for one TBRG. This action is then followed by assertion of the SDA pin (SDA = 0) for one TBRG while SCL is high. Following this, the RSEN bit (SSPCON2<1>) will be automatically cleared and the Baud Rate Generator will not be reloaded, leaving the SDA pin held low. As soon as a Start condition is detected on the SDA and SCL pins, the S bit (SSPSTAT<3>) will be set. The SSPIF bit will not be set until the Baud Rate Generator has timed out.

- Note 1: If RSEN is programmed while any other event is in progress, it will not take effect.
 - **2:** A bus collision during the Repeated Start condition occurs if:
 - SDA is sampled low when SCL goes from low-to-high.
 - SCL goes low before SDA is asserted low. This may indicate that another master is attempting to transmit a data '1'.

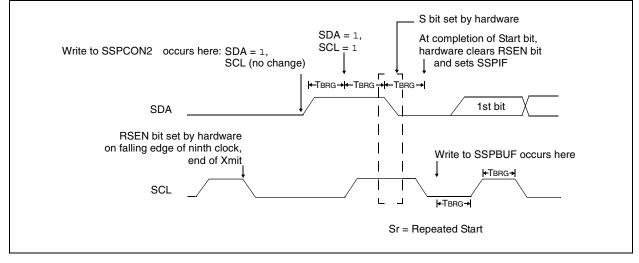
Immediately following the SSPIF bit getting set, the user may write the SSPBUF with the 7-bit address in 7-bit mode or the default first address in 10-bit mode. After the first eight bits are transmitted and an ACK is received, the user may then transmit an additional eight bits of address (10-bit mode) or eight bits of data (7-bit mode).

16.4.9.1 WCOL Status Flag

If the user writes the SSPBUF when a Repeated Start sequence is in progress, the WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).

Note: Because queueing of events is not allowed, writing of the lower 5 bits of SSPCON2 is disabled until the Repeated Start condition is complete.

FIGURE 16-22: REPEATED START CONDITION WAVEFORM



16.4.10 I²C MASTER MODE TRANSMISSION

Transmission of a data byte, a 7-bit address or the other half of a 10-bit address is accomplished by simply writing a value to the SSPBUF register. This action will set the Buffer Full bit, BF, and allow the Baud Rate Generator to begin counting and start the next transmission. Each bit of address/data will be shifted out onto the SDA pin after the falling edge of SCL is asserted (see data hold time specification parameter 106). SCL is held low for one Baud Rate Generator rollover count (TBRG). Data should be valid before SCL is released high (see data setup time specification parameter 107). When the SCL pin is released high, it is held that way for TBRG. The data on the SDA pin must remain stable for that duration and some hold time after the next falling edge of SCL. After the eighth bit is shifted out (the falling edge of the eighth clock), the BF flag is cleared and the master releases SDA. This allows the slave device being addressed to respond with an ACK bit during the ninth bit time if an address match occurred, or if data was received properly. The status of ACK is written into the ACKDT bit on the falling edge of the ninth clock. If the master receives an Acknowledge, the Acknowledge Status bit, ACKSTAT, is cleared; if not, the bit is set. After the ninth clock, the SSPIF bit is set and the master clock (Baud Rate Generator) is suspended until the next data byte is loaded into the SSPBUF, leaving SCL low and SDA unchanged (Figure 16-23).

After the write to the SSPBUF, each bit of the address will be shifted out on the falling edge of SCL until all seven address bits and the R/W bit are completed. On the falling edge of the eighth clock, the master will deassert the SDA pin, allowing the slave to respond with an Acknowledge. On the falling edge of the ninth clock, the master will sample the SDA pin to see if the address was recognized by a slave. The status of the ACK bit is loaded into the ACKSTAT status bit (SSPCON2<6>). Following the falling edge of the ninth clock transmission of the address, the SSPIF is set, the BF flag is cleared and the Baud Rate Generator is turned off until another write to the SSPBUF takes place, holding SCL low and allowing SDA to float.

16.4.10.1 BF Status Flag

In Transmit mode, the BF bit (SSPSTAT<0>) is set when the CPU writes to SSPBUF and is cleared when all 8 bits are shifted out.

16.4.10.2 WCOL Status Flag

If the user writes to the SSPBUF when a transmit is already in progress (i.e., SSPSR is still shifting out a data byte), the WCOL is set and the contents of the buffer are unchanged (the write doesn't occur) after 2 TcY after the SSPBUF write. If SSPBUF is rewritten within 2 TcY, the WCOL bit is set and SSPBUF is updated. This may result in a corrupted transfer. The user should verify that the WCOL is clear after each write to SSPBUF to ensure the transfer is correct. In all cases, WCOL must be cleared in software.

16.4.10.3 ACKSTAT Status Flag

In Transmit mode, the ACKSTAT bit (SSPCON2<6>) is cleared when the slave has sent an Acknowledge $(\overline{ACK} = 0)$ and is set when the slave does not Acknowledge $(\overline{ACK} = 1)$. A slave sends an Acknowledge when it has recognized its address (including a general call), or when the slave has properly received its data.

16.4.11 I²C MASTER MODE RECEPTION

Master mode reception is enabled by programming the Receive Enable bit, RCEN (SSPCON2<3>).

Note: The MSSP module must be in an Idle state before the RCEN bit is set or the RCEN bit will be disregarded.

The Baud Rate Generator begins counting and on each rollover, the state of the SCL pin changes (high-to-low/low-to-high) and data is shifted into the SSPSR. After the falling edge of the eighth clock, the receive enable flag is automatically cleared, the contents of the SSPSR are loaded into the SSPBUF, the BF flag bit is set, the SSPIF flag bit is set and the Baud Rate Generator is suspended from counting, holding SCL low. The MSSP is now in Idle state awaiting the next command. When the buffer is read by the CPU, the BF flag bit is automatically cleared. The user can then send an Acknowledge bit at the end of reception by setting the Acknowledge Sequence Enable bit, ACKEN (SSPCON2<4>).

16.4.11.1 BF Status Flag

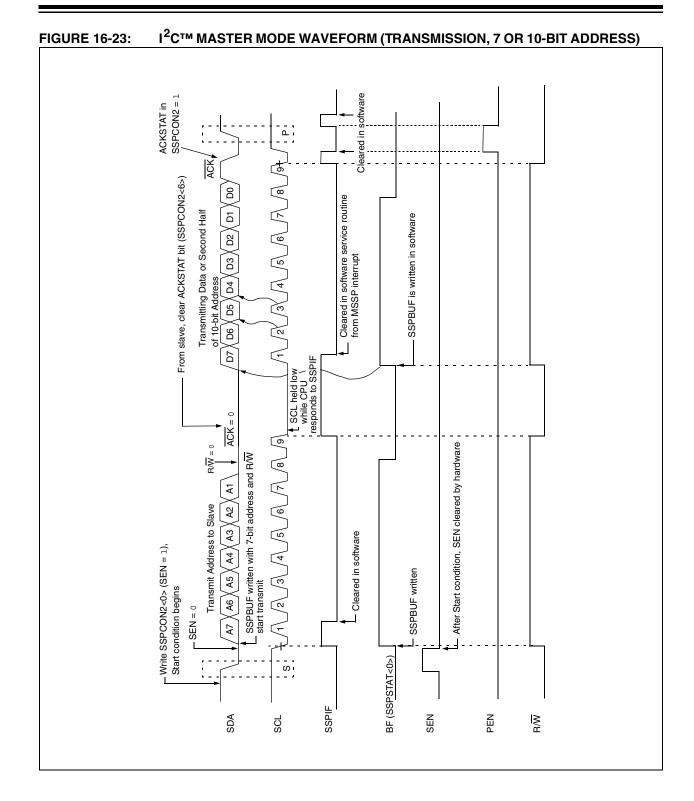
In receive operation, the BF bit is set when an address or data byte is loaded into SSPBUF from SSPSR. It is cleared when the SSPBUF register is read.

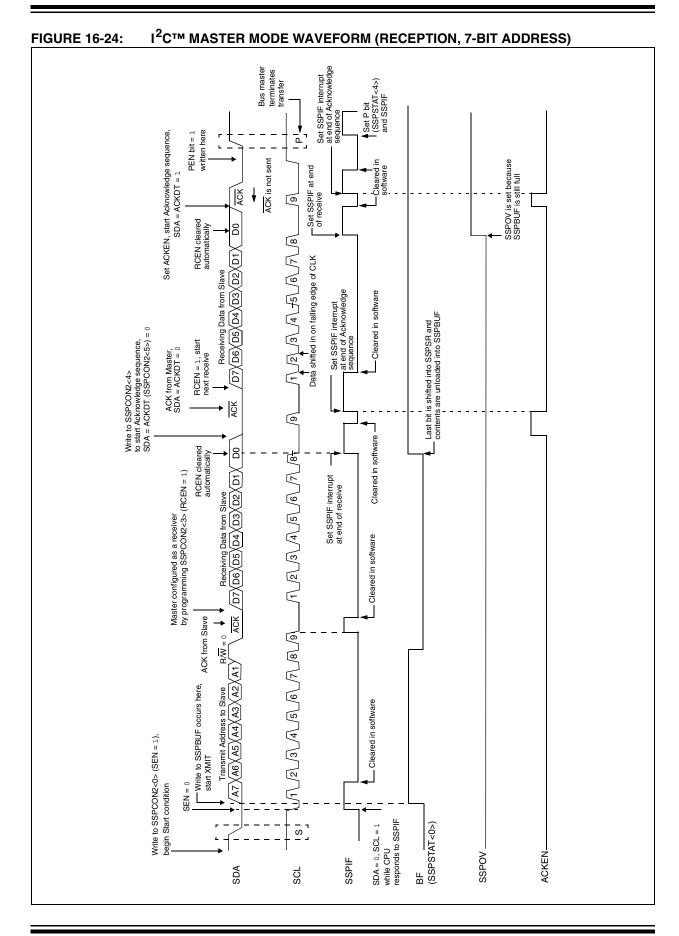
16.4.11.2 SSPOV Status Flag

In receive operation, the SSPOV bit is set when 8 bits are received into the SSPSR and the BF flag bit is already set from a previous reception.

16.4.11.3 WCOL Status Flag

If the user writes the SSPBUF when a receive is already in progress (i.e., SSPSR is still shifting in a data byte), the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur).





16.4.12 ACKNOWLEDGE SEQUENCE TIMING

An Acknowledge sequence is enabled by setting the Acknowledge Sequence Enable bit. ACKEN (SSPCON2<4>). When this bit is set, the SCL pin is pulled low and the contents of the Acknowledge data bit are presented on the SDA pin. If the user wishes to generate an Acknowledge, then the ACKDT bit should be cleared. If not, the user should set the ACKDT bit before starting an Acknowledge sequence. The Baud Rate Generator then counts for one rollover period (TBRG) and the SCL pin is deasserted (pulled high). When the SCL pin is sampled high (clock arbitration), the Baud Rate Generator counts for TBRG. The SCL pin is then pulled low. Following this, the ACKEN bit is automatically cleared, the Baud Rate Generator is turned off and the MSSP module then goes into Idle mode (Figure 16-25).

16.4.12.1 WCOL Status Flag

If the user writes the SSPBUF when an Acknowledge sequence is in progress, then WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).

16.4.13 STOP CONDITION TIMING

A Stop bit is asserted on the SDA pin at the end of a receive/transmit by setting the Stop Sequence Enable bit, PEN (SSPCON2<2>). At the end of a receive/transmit, the SCL line is held low after the falling edge of the ninth clock. When the PEN bit is set, the master will assert the SDA line low. When the SDA line is sampled low, the Baud Rate Generator is reloaded and counts down to '0'. When the Baud Rate Generator times out, the SCL pin will be brought high and one TBRG (Baud Rate Generator rollover count) later, the SDA pin will be deasserted. When the SDA pin is sampled high while SCL is high, the P bit (SSPSTAT<4>) is set. A TBRG later, the PEN bit is cleared and the SSPIF bit is set (Figure 16-26).

16.4.13.1 WCOL Status Flag

If the user writes the SSPBUF when a Stop sequence is in progress, then the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur).

FIGURE 16-25: ACKNOWLEDGE SEQUENCE WAVEFORM

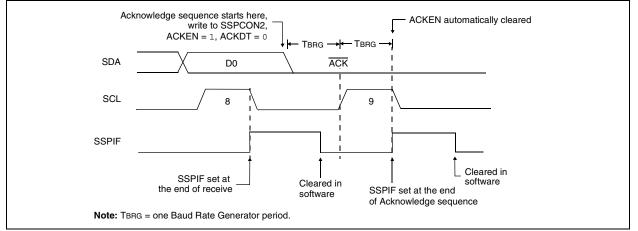
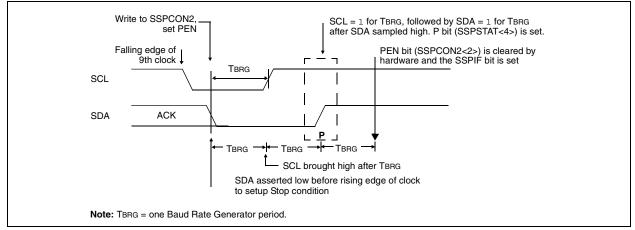


FIGURE 16-26: STOP CONDITION RECEIVE OR TRANSMIT MODE



16.4.14 SLEEP OPERATION

While in Sleep mode, the I^2C module can receive addresses or data and when an address match or complete byte transfer occurs, wake the processor from Sleep (if the MSSP interrupt is enabled).

16.4.15 EFFECTS OF A RESET

A Reset disables the MSSP module and terminates the current transfer.

16.4.16 MULTI-MASTER MODE

In Multi-Master mode, the interrupt generation on the detection of the Start and Stop conditions allows the determination of when the bus is free. The Stop (P) and Start (S) bits are cleared from a Reset or when the MSSP module is disabled. Control of the I^2C bus may be taken when the P bit (SSPSTAT<4>) is set, or the bus is Idle, with both the S and P bits clear. When the bus is busy, enabling the MSSP interrupt will generate the interrupt when the Stop condition occurs.

In multi-master operation, the SDA line must be monitored for arbitration to see if the signal level is the expected output level. This check is performed in hardware with the result placed in the BCLIF bit.

The states where arbitration can be lost are:

- · Address Transfer
- Data Transfer
- A Start Condition
- A Repeated Start Condition
- An Acknowledge Condition

16.4.17 MULTI -MASTER COMMUNICATION, BUS COLLISION AND BUS ARBITRATION

Multi-Master mode support is achieved by bus arbitration. When the master outputs address/data bits onto the SDA pin, arbitration takes place when the master outputs a '1' on SDA by letting SDA float high, and another master asserts a '0'. When the SCL pin floats high, data should be stable. If the expected data on SDA is a '1' and the data sampled on the SDA pin = 0, then a bus collision has taken place. The master will set the Bus Collision Interrupt Flag, BCLIF and reset the I^2C port to its Idle state (Figure 16-27).

If a transmit was in progress when the bus collision occurred, the transmission is halted, the BF flag is cleared, the SDA and SCL lines are deasserted and the SSPBUF can be written to. When the user services the bus collision Interrupt Service Routine and if the l^2C bus is free, the user can resume communication by asserting a Start condition.

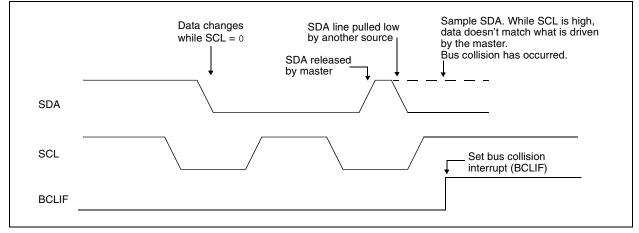
If a Start, Repeated Start, Stop or Acknowledge condition was in progress when the bus collision occurred, the condition is aborted, the SDA and SCL lines are deasserted and the respective control bits in the SSPCON2 register are cleared. When the user services the bus collision Interrupt Service Routine and if the l^2C bus is free, the user can resume communication by asserting a Start condition.

The master will continue to monitor the SDA and SCL pins. If a Stop condition occurs, the SSPIF bit will be set.

A write to the SSPBUF will start the transmission of data at the first data bit regardless of where the transmitter left off when the bus collision occurred.

In Multi-Master mode, the interrupt generation on the detection of Start and Stop conditions allows the determination of when the bus is free. Control of the I^2C bus can be taken when the P bit is set in the SSPSTAT register, or the bus is Idle and the S and P bits are cleared.

FIGURE 16-27: BUS COLLISION TIMING FOR TRANSMIT AND ACKNOWLEDGE



16.4.17.1 Bus Collision During a Start Condition

During a Start condition, a bus collision occurs if:

- a) SDA or SCL are sampled low at the beginning of the Start condition (Figure 16-28).
- b) SCL is sampled low before SDA is asserted low (Figure 16-29).

During a Start condition, both the SDA and the SCL pins are monitored.

If the SDA pin is already low, or the SCL pin is already low, then all of the following occur:

- the Start condition is aborted;
- · the BCLIF flag is set; and
- · the MSSP module is reset to its Idle state (Figure 16-28).

The Start condition begins with the SDA and SCL pins deasserted. When the SDA pin is sampled high, the Baud Rate Generator is loaded from SSPADD<6:0> and counts down to '0'. If the SCL pin is sampled low while SDA is high, a bus collision occurs, because it is assumed that another master is attempting to drive a data '1' during the Start condition.

If the SDA pin is sampled low during this count, the BRG is reset and the SDA line is asserted early (Figure 16-30). If, however, a '1' is sampled on the SDA pin, the SDA pin is asserted low at the end of the BRG count. The Baud Rate Generator is then reloaded and counts down to '0'. If the SCL pin is sampled as '0' during this time, a bus collision does not occur. At the end of the BRG count, the SCL pin is asserted low.

Note: The reason that bus collision is not a factor during a Start condition is that no two bus masters can assert a Start condition at the exact same time. Therefore, one master will always assert SDA before the other. This condition does not cause a bus collision because the two masters must be allowed to arbitrate the first address following the Start condition. If the address is the same, arbitration must be allowed to continue into the data portion, Repeated Start or Stop conditions.

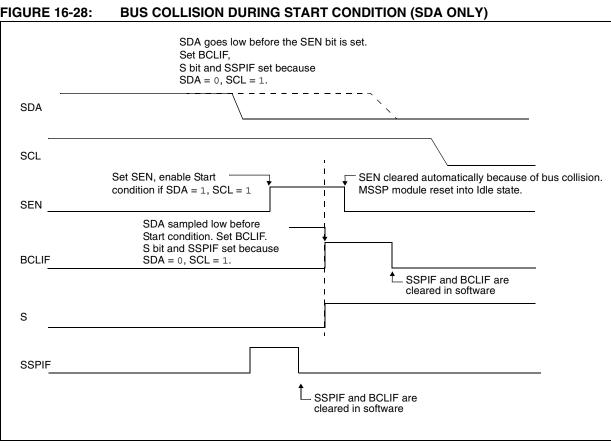


FIGURE 16-28:

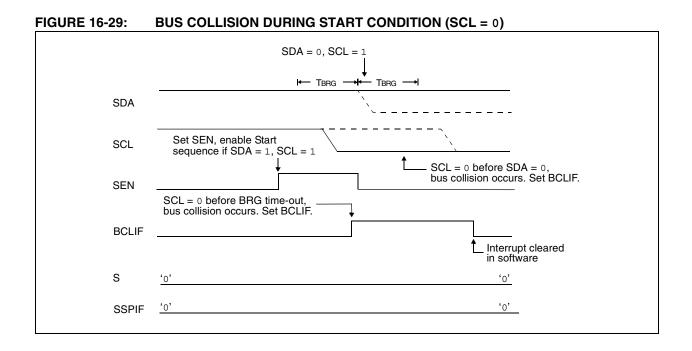
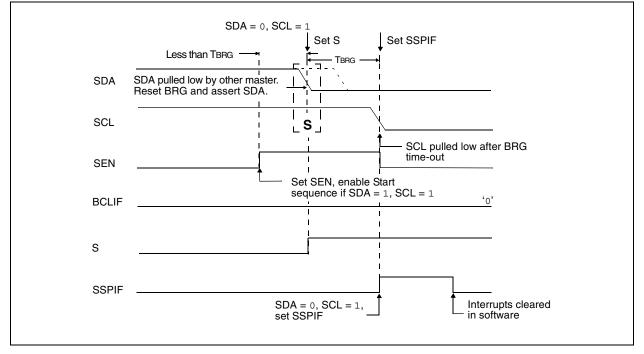


FIGURE 16-30: BRG RESET DUE TO SDA ARBITRATION DURING START CONDITION



16.4.17.2 Bus Collision During a Repeated Start Condition

During a Repeated Start condition, a bus collision occurs if:

- a) A low level is sampled on SDA when SCL goes from low level to high level.
- b) SCL goes low before SDA is asserted low, indicating that another master is attempting to transmit a data '1'.

When the user deasserts SDA and the pin is allowed to float high, the BRG is loaded with SSPADD<6:0> and counts down to '0'. The SCL pin is then deasserted and when sampled high, the SDA pin is sampled.

If SDA is low, a bus collision has occurred (i.e., another master is attempting to transmit a data '0', see Figure 16-31). If SDA is sampled high, the BRG is reloaded and begins counting. If SDA goes from high-to-low before the BRG times out, no bus collision occurs because no two masters can assert SDA at exactly the same time.

If SCL goes from high-to-low before the BRG times out and SDA has not already been asserted, a bus collision occurs. In this case, another master is attempting to transmit a data '1' during the Repeated Start condition (see Figure 16-32).

If, at the end of the BRG time-out, both SCL and SDA are still high, the SDA pin is driven low and the BRG is reloaded and begins counting. At the end of the count, regardless of the status of the SCL pin, the SCL pin is driven low and the Repeated Start condition is complete.



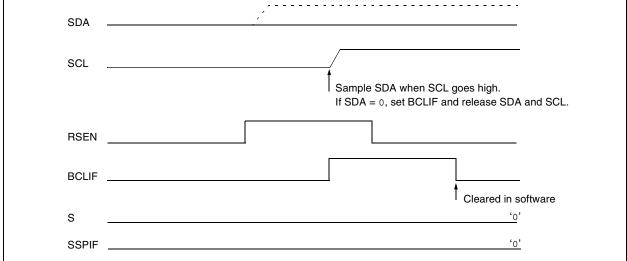
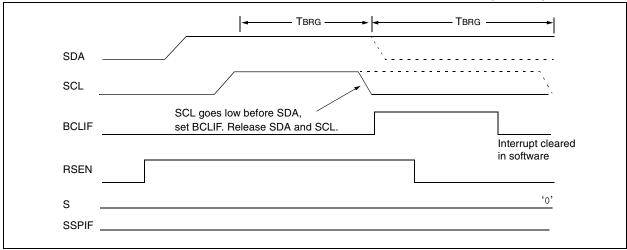


FIGURE 16-32: BUS COLLISION DURING REPEATED START CONDITION (CASE 2)



16.4.17.3 Bus Collision During a Stop Condition

Bus collision occurs during a Stop condition if:

- a) After the SDA pin has been deasserted and allowed to float high, SDA is sampled low after the BRG has timed out.
- b) After the SCL pin is deasserted, SCL is sampled low before SDA goes high.

The Stop condition begins with SDA asserted low. When SDA is sampled low, the SCL pin is allowed to float. When the pin is sampled high (clock arbitration), the Baud Rate Generator is loaded with SSPADD<6:0> and counts down to '0'. After the BRG times out, SDA is sampled. If SDA is sampled low, a bus collision has occurred. This is due to another master attempting to drive a data '0' (Figure 16-33). If the SCL pin is sampled low before SDA is allowed to float high, a bus collision occurs. This is another case of another master attempting to drive a data '0' (Figure 16-34).

FIGURE 16-33: BUS COLLISION DURING A STOP CONDITION (CASE 1)

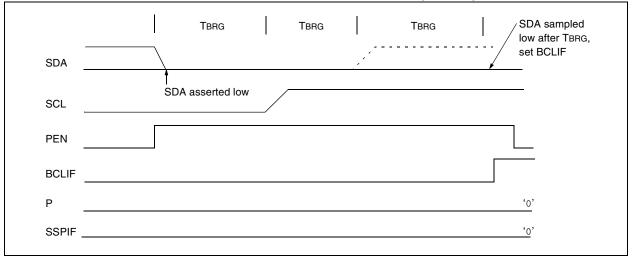
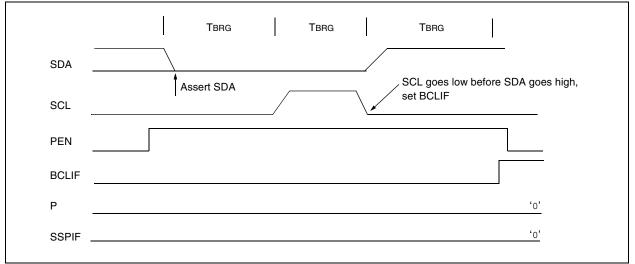


FIGURE 16-34: BUS COLLISION DURING A STOP CONDITION (CASE 2)



Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51
PIR1	—	ADIF	RC1IF	TX1IF	SSPIF	—	TMR2IF	TMR1IF	54
PIE1	—	ADIE	RC1IE	TX1IE	SSPIE	—	TMR2IE	TMR1IE	54
IPR1	—	ADIP	RC1IP	TX1IP	SSPIP	—	TMR2IP	TMR1IP	54
PIR2	OSCFIF	CMIF	—	—	BCLIF	LVDIF	TMR3IF		54
PIE2	OSCFIE	CMIE	—	—	BCLIE	LVDIE	TMR3IE	_	54
IPR2	OSCFIP	CMIP	—	—	BCLIP	LVDIP	TMR3IP	_	54
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	54
SSPBUF	MSSP Rec	eive Buffer/7	Fransmit Reg	gister					52
SSPADD		ress Registe d Rate Relo			mode)				52
SSPCON1	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	52
SSPCON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	52
	GCEN	ACKSTAT	ADMSK5 ⁽¹⁾	ADMSK4 ⁽¹⁾	ADMSK3 ⁽¹⁾	ADMSK2(1)	ADMSK1 ⁽¹⁾	SEN	
SSPSTAT	SMP	CKE	D/Ā	Р	S	R/W	UA	BF	52
			(1 0)						

TABLE 16-4: REGISTERS ASSOCIATED WITH I²C[™] OPERATION

Legend: -= unimplemented, read as '0'. Shaded cells are not used by the MSSP module in I²CTM mode.

Note 1: Alternate bit definitions for use in I²C Slave mode operations only.

17.0 ENHANCED UNIVERSAL SYNCHRONOUS ASYNCHRONOUS RECEIVER TRANSMITTER (EUSART)

PIC18F85J90 family devices have three serial I/O modules: the MSSP module, discussed in the previous chapter and two Universal Synchronous Asynchronous Receiver Transmitter (USART) modules. (Generically, the USART is also known as a Serial Communications Interface or SCI.) The USART can be configured as a full-duplex asynchronous system that can communicate with peripheral devices, such as CRT terminals and personal computers. It can also be configured as a half-duplex synchronous system that can communicate with peripheral devices, such as A/D or D/A integrated circuits, serial EEPROMs, etc.

There are two distinct implementations of the USART module in these devices: the Enhanced USART (EUSART) discussed here and the Addressable USART discussed in the next chapter. For this device family, USART1 always refers to the EUSART, while USART2 is always the AUSART.

The EUSART and AUSART modules implement the same core features for serial communications; their basic operation is essentially the same. The EUSART module provides additional features, including Automatic Baud Rate Detection and calibration, automatic wake-up on Sync Break reception and 12-bit Break character transmit. These features make it ideally suited for use in Local Interconnect Network bus (LIN bus) systems.

The EUSART can be configured in the following modes:

- Asynchronous (full-duplex) with:
 - Auto-wake-up on character reception
 - Auto-baud calibration
 - 12-bit Break character transmission
- Synchronous Master (half-duplex) with selectable clock polarity
- Synchronous Slave (half-duplex) with selectable clock polarity

The pins of the EUSART are multiplexed with the functions of PORTC (RC6/TX1/CK1/SEG27 and RC7/RX1/DT1/SEG28). In order to configure these pins as an EUSART:

- bit SPEN (RCSTA1<7>) must be set (= 1)
- bit TRISC<7> must be set (= 1)
- bit TRISC<6> must be set (= 1)
- Note: The EUSART control will automatically reconfigure the pin from input to output as needed.

The driver for the TX1 output pin can also be optionally configured as an open-drain output. This feature allows the voltage level on the pin to be pulled to a higher level through an external pull-up resistor, and allows the output to communicate with external circuits without the need for additional level shifters.

The open-drain output option is controlled by the U1OD bit (LATG<6>). Setting the bit configures the pin for open-drain operation.

17.1 Control Registers

The operation of the Enhanced USART module is controlled through three registers:

- Transmit Status and Control Register 1 (TXSTA1)
- Receive Status and Control Register 1 (RCSTA1)
- Baud Rate Control Register 1 (BAUDCON1)

The registers are described in Register 17-1, Register 17-2 and Register 17-3.

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-1	R/W-0			
CSRC	TX9	TXEN ⁽¹⁾	SYNC	SENDB	BRGH	TRMT	TX9D			
bit 7							bit			
Legend:										
R = Readab		W = Writable		-	nented bit, rea					
-n = Value a	t POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown			
bit 7	CSRC: Clock	Source Select	bit							
	<u>Asynchronou</u> Don't care.	<u>is mode:</u>								
		<u>s mode:</u> node (clock gen node (clock from								
bit 6	TX9: 9-Bit Tr	ansmit Enable I	bit							
		9-bit transmissio 3-bit transmissio								
oit 5	TXEN: Trans	mit Enable bit ⁽¹)							
	1 = Transmit 0 = Transmit									
bit 4	SYNC: AUSART Mode Select bit									
	1 = Synchro 0 = Asynchro									
bit 3	SENDB: Ser	d Break Chara	cter bit							
	<u>Asynchronou</u>	<u>is mode:</u>								
		nc Break on ne eak transmissio		n (cleared by ha	ardware upon o	completion)				
	<u>Synchronous</u> Don't care.	mode:								
bit 2	BRGH: High	Baud Rate Sel	ect bit							
	Asynchronou 1 = High spe 0 = Low spe	ed								
	Synchronous Unused in thi	mode:								
bit 1		mit Shift Regist	er Status bit							
	1 = TSR em 0 = TSR full	•								
bit 0	TX9D: 9th bi	t of Transmit Da	ata							
	Can be addre	ess/data bit or a	a parity bit.							

REGISTER 17-1: TXSTA1: EUSART TRANSMIT STATUS AND CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-0	R-0	R-x
SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D
oit 7							bit (
_egend:							
R = Readab		W = Writable b	oit	U = Unimplem			
-n = Value a	t POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkn	own
bit 7	SPEN: Seria	l Port Enable bit					
		ort enabled (conf	igures RX1/D	T1 and TX1/CK	1 pins as seri	al port pins)	
		ort disabled (held					
bit 6	RX9: 9-Bit R	eceive Enable bi	t				
		9-bit reception 8-bit reception					
bit 5	SREN: Singl	e Receive Enabl	e bit				
	<u>Asynchronou</u> Don't care.	<u>is mode</u> :					
		s mode – Master	<u>.</u>				
	1 = Enables	single receive					
		s single receive ared after recep	tion is comple	ete.			
	<u>Synchronous</u> Don't care.	<u>s mode – Slave:</u>					
bit 4	CREN: Cont	inuous Receive I	Enable bit				
	Asynchronou 1 = Enables 0 = Disables	receiver					
		s mode: continuous rece s continuous rece		ole bit CREN is c	leared (CREN	V overrides SRE	N)
oit 3		dress Detect Ena					
bit 0		us mode 9-bit (R)					
	1 = Enables	address detection	on, enables ir				
		address detecti	-	are received and	d ninth bit can	be used as pari	ty bit
	<u>Asynchronou</u> Don't care.	<u>is mode 9-bit (R)</u>	<u> (9 = 0)</u> :				
bit 2	FERR: Fram	ing Error bit					
		error (can be up	dated by rea	ding BCBEG1 re	egister and re	ceiving next vali	d byte)
	0 = No fram	· ·			ogiotor and ro		a byto,
oit 1	OERR: Over	run Error bit					
	1 = Overrun 0 = No over	error (can be cle run error	eared by clea	ring bit CREN)			
oit 0	RX9D: 9th bi	it of Received Da	ata				
	This says have	address/data bit					

R/W-0	R-1	U-0	R/W-0	R/W-0	U-0	R/W-0	R/W-0
ABDOVF	RCMT	_	SCKP	BRG16	_	WUE	ABDEN
bit 7							bit (
Legend:							
R = Readabl	le bit	W = Writable	bit	U = Unimplem	ented bit, read	d as '0'	
-n = Value at	t POR	'1' = Bit is set	t	'0' = Bit is clea	red	x = Bit is unkr	nown
bit 7		uto-Baud Acqui					
		rollover has occ e cleared in sof	-	uto-Baud Rate D	Detect mode		
	`	rollover has o	,				
bit 6	RCMT: Rece	ive Operation I	dle Status bit				
		operation is Id					
	0 = Receive	operation is ac	tive				
bit 5	•	nted: Read as					
bit 4	-	hronous Clock	Polarity Select	bit			
	Asynchronou Unused in th						
	Synchronous						
		e for clock (CK ⁻	1) is a high leve	əl			
		e for clock (CK	, .				
bit 3	BRG16: 16-I	Bit Baud Rate F	Register Enable	e bit			
				H1 and SPBRG			
				only (Compatib	le mode), SPE	BRGH1 value ig	gnored
bit 2	-	nted: Read as	'0'				
bit 1		-up Enable bit					
	Asynchronou		to sample the l	RX1 pin – interru	int concrated	on falling edge	, bit cleared in
		e on following r			ipi generaleu	on failing edge	, bit cleared i
		not monitored		detected			
	Synchronous						
	Unused in th						
bit 0		o-Baud Detect	Enable bit				
	<u>Asynchronou</u> 1 – Enable		surement on th	e next characte	r Roquiros ro	cention of a Sv	inc field (55h)
		in hardware up			. riequires re		
		te measuremer					
	Synchronous						
	Unused in th	is mode.					

17.2 **EUSART Baud Rate Generator** (BRG)

The BRG is a dedicated, 8-bit or 16-bit generator that supports both the Asynchronous and Synchronous modes of the EUSART. By default, the BRG operates in 8-bit mode; setting the BRG16 bit (BAUDCON1<3>) selects 16-bit mode.

The SPBRGH1:SPBRG1 register pair controls the period of a free-running timer. In Asynchronous mode, BRGH (TXSTA1<2>) and BRG16 (BAUDCON1<3>) bits also control the baud rate. In Synchronous mode, BRGH is ignored. Table 17-1 shows the formula for computation of the baud rate for different EUSART modes that only apply in Master mode (internally generated clock).

Given the desired baud rate and FOSC, the nearest integer value for the SPBRGH1:SPBRG1 registers can be calculated using the formulas in Table 17-1. From this, the error in baud rate can be determined. An example calculation is shown in Example 17-1. Typical baud rates and error values for the various Asynchronous modes are shown in Table 17-2. It may be advantageous to use the high baud rate (BRGH = 1) or the 16-bit BRG to reduce the baud rate error, or achieve a slow baud rate for a fast oscillator frequency.

Writing a new value to the SPBRGH1:SPBRG1 registers causes the BRG timer to be reset (or cleared). This ensures the BRG does not wait for a timer overflow before outputting the new baud rate.

17.2.1 **OPERATION IN POWER-MANAGED** MODES

The device clock is used to generate the desired baud rate. When one of the power-managed modes is entered, the new clock source may be operating at a different frequency. This may require an adjustment to the value in the SPBRG1 register pair.

SAMPLING 17.2.2

The data on the RX1 pin is sampled three times by a majority detect circuit to determine if a high or a low level is present at the RX1 pin.

_AS
BBG/EUSAB
-

Co	onfiguration B	its		Baud Rate Formula		
SYNC	BRG16	BRGH	BRG/EUSART Mode	Baud Rate Formula		
0	0	0	8-bit/Asynchronous	Fosc/[64 (n + 1)]		
0	0	1	8-bit/Asynchronous			
0	1	0	16-bit/Asynchronous	Fosc/[16 (n + 1)]		
0	1	1	16-bit/Asynchronous			
1	0	x	8-bit/Synchronous	Fosc/[4 (n + 1)]		
1	1	x	16-bit/Synchronous			

Legend: x = Don't care, n = Value of SPBRGH1:SPBRG1 register pair

EXAMPLE 17-1: CALCULATING BAUD RATE ERROR

For a device with FOSC of 16 MHz, desired baud rate of 9600, Asynchronous mode, 8-bit BRG:
Desired Baud Rate = FOSC/(64 ([SPBRGH1:SPBRG1] + 1))
Solving for SPBRGH1:SPBRG1:
X = ((Fosc/Desired Baud Rate)/64) - 1
= ((1600000/9600)/64) - 1
= [25.042] = 25
Calculated Baud Rate = $16000000/(64(25+1))$
= 9615
Error = (Calculated Baud Rate – Desired Baud Rate)/Desired Baud Rate
= (9615 - 9600)/9600 = 0.16%

TABLE 17-2: **REGISTERS ASSOCIATED WITH THE BAUD RATE GENERATOR**

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page	
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	53	
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	53	
BAUDCON1	ABDOVF	RCMT	—	SCKP	BRG16	—	WUE	ABDEN	55	
SPBRGH1	EUSART E	Baud Rate	Generator	r Register	High Byte				55	
SPBRG1		53								
SPBRG1	SPBRG1 EUSART Baud Rate Generator Register Low Byte									

Legend: - = unimplemented, read as '0'. Shaded cells are not used by the BRG.

					SYNC	= 0, BRGH	l = 0, BRG	G16 = 0				
BAUD RATE	Fosc	= 40.000) MHz	Fosc = 20.000 MHz			Fosc = 10.000 MHz			Fosc = 8.000 MHz		
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	—	_	_	_	_	_	_	_	_	-	_	_
1.2	—	—	—	1.221	1.73	255	1.202	0.16	129	1.201	-0.16	103
2.4	2.441	1.73	255	2.404	0.16	129	2.404	0.16	64	2.403	-0.16	51
9.6	9.615	0.16	64	9.766	1.73	31	9.766	1.73	15	9.615	-0.16	12
19.2	19.531	1.73	31	19.531	1.73	15	19.531	1.73	7	—	_	_
57.6	56.818	-1.36	10	62.500	8.51	4	52.083	-9.58	2	—	_	_
115.2	125.000	8.51	4	104.167	-9.58	2	78.125	-32.18	1	—	—	—

TABLE 17-3: BAUD RATES FOR ASYNCHRONOUS MODES

			S	YNC = 0, E	BRGH = (, BRG16 =	0				
BAUD	Fos	c = 4.000	MHz	Fos	c = 2.000	MHz	Fos	Fosc = 1.000 MHz			
RATE (K)	Actual Rate (K)	Rate Error value		Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)		
0.3	0.300	0.16	207	0.300	-0.16	103	0.300	-0.16	51		
1.2	1.202	0.16	51	1.201	-0.16	25	1.201	-0.16	12		
2.4	2.404	0.16	25	2.403	-0.16	12	—	—	_		
9.6	8.929	-6.99	6	—	_	—	_	_	—		
19.2	20.833	8.51	2	—	_	—	_	_	—		
57.6	62.500	8.51	0	—	_	_	—	_	_		
115.2	62.500	-45.75	0	—	—	—	_	—	—		

					SYNC	= 0, BRGH	i = 1, BRG	i 16 = 0				
BAUD RATE	Fosc	Fosc = 40.000 MHz Fosc = 20.00			= 20.000	MHz Fosc = 10.000 MHz				Fosc = 8.000 MHz		
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	—	_	_	_	_	_	_	_	_	_	_	_
1.2	—	_	—	—	—	—	—	_	—	—	_	—
2.4	—	—	—	—	—	—	2.441	1.73	255	2.403	-0.16	207
9.6	9.766	1.73	255	9.615	0.16	129	9.615	0.16	64	9.615	-0.16	51
19.2	19.231	0.16	129	19.231	0.16	64	19.531	1.73	31	19.230	-0.16	25
57.6	58.140	0.94	42	56.818	-1.36	21	56.818	-1.36	10	55.555	3.55	8
115.2	113.636	-1.36	21	113.636	-1.36	10	125.000	8.51	4	_		—

			S	YNC = 0, E	BRGH = 1	L, BRG16 =	0			
BAUD	Foso	c = 4.000	MHz	Fos	c = 2.000	MHz	Fosc = 1.000 MHz			
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	
0.3	_	_	_	_	_	_	0.300	-0.16	207	
1.2	1.202	0.16	207	1.201	-0.16	103	1.201	-0.16	51	
2.4	2.404	0.16	103	2.403	-0.16	51	2.403	-0.16	25	
9.6	9.615	0.16	25	9.615	-0.16	12	—	—	—	
19.2	19.231	0.16	12	_	_	—	_	_	—	
57.6	62.500	8.51	3	—	_	—	—	—	—	
115.2	125.000	8.51	1	_	—	—	_	—	—	

		SYNC = 0, BRGH = 0, BRG16 = 1													
BAUD	Fosc	= 40.000) MHz	Fosc = 20.000 MHz			Fosc	= 10.000) MHz	Fos	Fosc = 8.000 MHz				
RATE (K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)			
0.3	0.300	0.00	8332	0.300	0.02	4165	0.300	0.02	2082	0.300	-0.04	1665			
1.2	1.200	0.02	2082	1.200	-0.03	1041	1.200	-0.03	520	1.201	-0.16	415			
2.4	2.402	0.06	1040	2.399	-0.03	520	2.404	0.16	259	2.403	-0.16	207			
9.6	9.615	0.16	259	9.615	0.16	129	9.615	0.16	64	9.615	-0.16	51			
19.2	19.231	0.16	129	19.231	0.16	64	19.531	1.73	31	19.230	-0.16	25			
57.6	58.140	0.94	42	56.818	-1.36	21	56.818	-1.36	10	55.555	3.55	8			
115.2	113.636	-1.36	21	113.636	-1.36	10	125.000	8.51	4	—	_	_			

TABLE 17-3: BAUD RATES FOR ASYNCHRONOUS MODES (CONTINUED)

			S	YNC = 0, E	BRGH = (, BRG16 =	1			
BAUD RATE	Foso	= 4.000	MHz	Fos	c = 2.000	MHz	Fosc = 1.000 MHz			
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	
0.3	0.300	0.04	832	0.300	-0.16	415	0.300	-0.16	207	
1.2	1.202	0.16	207	1.201	-0.16	103	1.201	-0.16	51	
2.4	2.404	0.16	103	2.403	-0.16	51	2.403	-0.16	25	
9.6	9.615	0.16	25	9.615	-0.16	12	—	_	—	
19.2	19.231	0.16	12	—	_	_	—	_	_	
57.6	62.500	8.51	3	—	_	_	—	_	_	
115.2	125.000	8.51	1	_		—	_	_		

				SYNC = 0	, BRGH =	= 1, BRG16	6 = 1 or SY	'NC = 1,	BRG16 = 1			
BAUD RATE	Fosc	= 40.000) MHz	Fosc	c = 20.000 MHz Fosc = 10.000 MHz) MHz	Fosc = 8.000 MHz			
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	0.300	0.00	33332	0.300	0.00	16665	0.300	0.00	8332	0.300	-0.01	6665
1.2	1.200	0.00	8332	1.200	0.02	4165	1.200	0.02	2082	1.200	-0.04	1665
2.4	2.400	0.02	4165	2.400	0.02	2082	2.402	0.06	1040	2.400	-0.04	832
9.6	9.606	0.06	1040	9.596	-0.03	520	9.615	0.16	259	9.615	-0.16	207
19.2	19.193	-0.03	520	19.231	0.16	259	19.231	0.16	129	19.230	-0.16	103
57.6	57.803	0.35	172	57.471	-0.22	86	58.140	0.94	42	57.142	0.79	34
115.2	114.943	-0.22	86	116.279	0.94	42	113.636	-1.36	21	117.647	-2.12	16

		SYN	IC = 0, BR(GH = 1, BI	RG16 = 1	or SYNC =	= 1, BRG1	6 = 1	
BAUD	Fos	c = 4.000	MHz	Fos	c = 2.000	MHz	Fosc = 1.000 MHz		
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	0.300	0.01	3332	0.300	-0.04	1665	0.300	-0.04	832
1.2	1.200	0.04	832	1.201	-0.16	415	1.201	-0.16	207
2.4	2.404	0.16	415	2.403	-0.16	207	2.403	-0.16	103
9.6	9.615	0.16	103	9.615	-0.16	51	9.615	-0.16	25
19.2	19.231	0.16	51	19.230	-0.16	25	19.230	-0.16	12
57.6	58.824	2.12	16	55.555	3.55	8	—		—
115.2	111.111	-3.55	8	_	—	—	—	—	—

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17.2.3 AUTO-BAUD RATE DETECT

The Enhanced USART module supports the automatic detection and calibration of baud rate. This feature is active only in Asynchronous mode and while the WUE bit is clear.

The automatic baud rate measurement sequence (Figure 17-1) begins whenever a Start bit is received and the ABDEN bit is set. The calculation is self-averaging.

In the Auto-Baud Rate Detect (ABD) mode, the clock to the BRG is reversed. Rather than the BRG clocking the incoming RX1 signal, the RX1 signal is timing the BRG. In ABD mode, the internal Baud Rate Generator is used as a counter to time the bit period of the incoming serial byte stream.

Once the ABDEN bit is set, the state machine will clear the BRG and look for a Start bit. The Auto-Baud Rate Detect must receive a byte with the value, 55h (ASCII "U", which is also the LIN bus Sync character), in order to calculate the proper bit rate. The measurement is taken over both a low and a high bit time in order to minimize any effects caused by asymmetry of the incoming signal. After a Start bit, the SPBRG1 begins counting up, using the preselected clock source on the first rising edge of RX1. After eight bits on the RX1 pin or the fifth rising edge, an accumulated value totalling the proper BRG period is left in the SPBRGH1:SPBRG1 register pair. Once the 5th edge is seen (this should correspond to the Stop bit), the ABDEN bit is automatically cleared.

If a rollover of the BRG occurs (an overflow from FFFFh to 0000h), the event is trapped by the ABDOVF status bit (BAUDCON1<7>). It is set in hardware by BRG rollovers and can be set or cleared by the user in software. ABD mode remains active after rollover events and the ABDEN bit remains set (Figure 17-2).

While calibrating the baud rate period, the BRG registers are clocked at 1/8th the preconfigured clock rate. Note that the BRG clock will be configured by the BRG16 and BRGH bits. Independent of the BRG16 bit setting, both the SPBRG1 and SPBRGH1 will be used as a 16-bit counter. This allows the user to verify that no carry occurred for 8-bit modes by checking for 00h in the SPBRGH1 register. Refer to Table 17-4 for counter clock rates to the BRG. While the ABD sequence takes place, the EUSART state machine is held in Idle. The RC1IF interrupt is set once the fifth rising edge on RX1 is detected. The value in the RCREG1 needs to be read to clear the RC1IF interrupt. The contents of RCREG1 should be discarded.

- Note 1: If the WUE bit is set with the ABDEN bit, Auto-Baud Rate Detection will occur on the byte *following* the Break character.
 - 2: It is up to the user to determine that the incoming character baud rate is within the range of the selected BRG clock source. Some combinations of oscillator frequency and EUSART baud rates are not possible due to bit error rates. Overall system timing and communication baud rates must be taken into consideration when using the Auto-Baud Rate Detection feature.

TABLE 17-4: BRG COUNTER CLOCK RATES

BRG16	BRGH	BRG Counter Clock
0	0	Fosc/512
0	1	Fosc/128
1	0	Fosc/128
1	1	Fosc/32

Note: During the ABD sequence, SPBRG1 and SPBRGH1 are both used as a 16-bit counter, independent of the BRG16 setting.

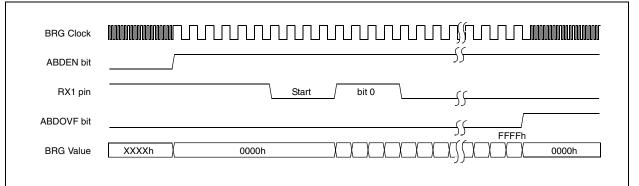
17.2.3.1 ABD and EUSART Transmission

Since the BRG clock is reversed during ABD acquisition, the EUSART transmitter cannot be used during ABD. This means that whenever the ABDEN bit is set, TXREG1 cannot be written to. Users should also ensure that ABDEN does not become set during a transmit sequence. Failing to do this may result in unpredictable EUSART operation.

			NATE CALCOLATION	
BRG Value	XXXXh	X 0000h (001Ch
RX1 pin		Start	Edge #1 Edge #2 Edge #3 Edge #4 bit 0 bit 1 bit 2 bit 3 bit 4 bit 5 bit 6 bit 7	← Edge #5 Stop bit
BRG Clock			www.ww	תתקתתתתתתתתתתתחתת
ABDEN bit	Set by User			Auto-Cleared
RC1IF bit (Interrupt)				
Read RCREG1		 		
SPBRG1		 	XXXXh	1Ch
SPBRGH1			xxxxh X	00h

FIGURE 17-1: AUTOMATIC BAUD RATE CALCULATION

FIGURE 17-2: BRG OVERFLOW SEQUENCE



17.3 EUSART Asynchronous Mode

The Asynchronous mode of operation is selected by clearing the SYNC bit (TXSTA1<4>). In this mode, the EUSART uses standard Non-Return-to-Zero (NRZ) format (one Start bit, eight or nine data bits and one Stop bit). The most common data format is 8 bits. An on-chip dedicated 8-bit/16-bit Baud Rate Generator can be used to derive standard baud rate frequencies from the oscillator.

The EUSART transmits and receives the LSb first. The EUSART's transmitter and receiver are functionally independent, but use the same data format and baud rate. The Baud Rate Generator produces a clock, either x16 or x64 of the bit shift rate, depending on the BRGH and BRG16 bits (TXSTA1<2> and BAUDCON1<3>). Parity is not supported by the hardware but can be implemented in software and stored as the 9th data bit.

When operating in Asynchronous mode, the EUSART module consists of the following important elements:

- Baud Rate Generator
- Sampling Circuit
- Asynchronous Transmitter
- Asynchronous Receiver
- Auto-Wake-up on Sync Break Character
- 12-Bit Break Character Transmit
- Auto-Baud Rate Detection

17.3.1 EUSART ASYNCHRONOUS TRANSMITTER

The EUSART transmitter block diagram is shown in Figure 17-3. The heart of the transmitter is the Transmit (Serial) Shift register (TSR). The Shift register obtains its data from the Read/Write Transmit Buffer register, TXREG1. The TXREG1 register is loaded with data in software. The TSR register is not loaded until the Stop bit has been transmitted from the previous load. As soon as the Stop bit is transmitted, the TSR is loaded with new data from the TXREG1 register (if available).

Once the TXREG1 register transfers the data to the TSR register (occurs in one TCY), the TXREG1 register is empty and the TX1IF flag bit (PIR1<4>) is set. This interrupt can be enabled or disabled by setting or clearing the interrupt enable bit, TX1IE (PIE1<4>). TX1IF will be set regardless of the state of TX1IE; it cannot be cleared in software. TX1IF is also not cleared immediately upon loading TXREG1, but becomes valid in the second instruction cycle following the load instruction. Polling TX1IF immediately following a load of TXREG1 will return invalid results.

While TX1IF indicates the status of the TXREG1 register, another bit, TRMT (TXSTA1<1>), shows the status of the TSR register. TRMT is a read-only bit which is set when the TSR register is empty. No interrupt logic is tied to this bit so the user has to poll this bit in order to determine if the TSR register is empty.

Note 1:	The TSR register is not mapped in data
	memory so it is not available to the user.
о.	Elag hit TV11E is set when enable hit

2: Flag bit, TX1IF, is set when enable bit, TXEN, is set.

To set up an Asynchronous Transmission:

- 1. Initialize the SPBRGH1:SPBRG1 registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
- 2. Enable the asynchronous serial port by clearing bit SYNC and setting bit SPEN.
- 3. If interrupts are desired, set enable bit, TX1IE.
- 4. If 9-bit transmission is desired, set transmit bit, TX9; can be used as address/data bit.
- 5. Enable the transmission by setting bit, TXEN, which will also set bit, TX1IF.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit TX9D.
- 7. Load data to the TXREG1 register (starts transmission).
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

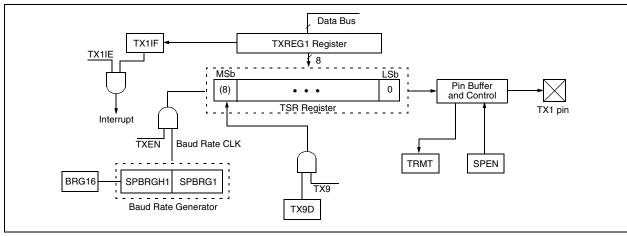


FIGURE 17-3: EUSART TRANSMIT BLOCK DIAGRAM

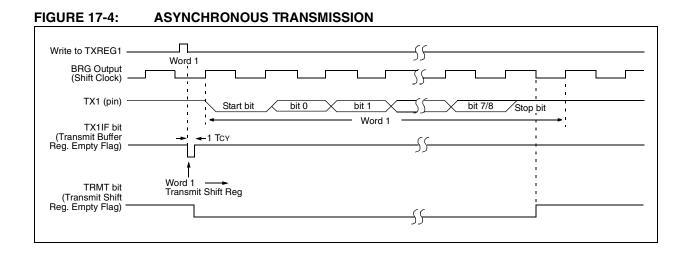


FIGURE 17-5: ASYNCHRONOUS TRANSMISSION (BACK TO BACK)

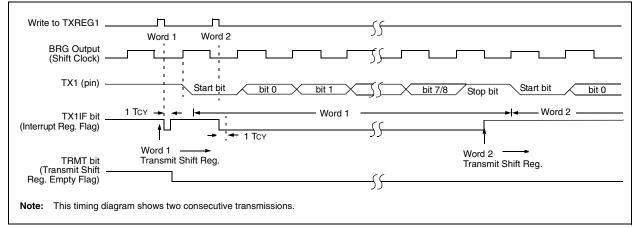


TABLE 17-5: REGISTERS ASSOCIATED WITH ASYNCHRONOUS TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51
PIR1	—	ADIF	RC1IF	TX1IF	SSPIF	—	TMR2IF	TMR1IF	54
PIE1	—	ADIE	RC1IE	TX1IE	SSPIE	_	TMR2IE	TMR1IE	54
IPR1	—	ADIP	RC1IP	TX1IP	SSPIP	_	TMR2IP	TMR1IP	54
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	53
TXREG1	EUSART T	ransmit Reg	ister						53
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	53
BAUDCON1	ABDOVF	RCMT	—	SCKP	BRG16	—	WUE	ABDEN	55
SPBRGH1	EUSART E	aud Rate G	enerator Re	gister High	Byte				55
SPBRG1	EUSART E	aud Rate G	enerator Re	gister Low	Byte				53
LATG	U2OD	U10D	_	LATG4	LATG3	LATG2	LATG1	LATG0	54

Legend: — = unimplemented locations read as '0'. Shaded cells are not used for asynchronous transmission.

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17.3.2 EUSART ASYNCHRONOUS RECEIVER

The receiver block diagram is shown in Figure 17-6. The data is received on the RX1 pin and drives the data recovery block. The data recovery block is actually a high-speed shifter operating at x16 times the baud rate, whereas the main receive serial shifter operates at the bit rate or at Fosc. This mode would typically be used in RS-232 systems.

To set up an Asynchronous Reception:

- 1. Initialize the SPBRGH1:SPBRG1 registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
- 2. Enable the asynchronous serial port by clearing bit SYNC and setting bit, SPEN.
- 3. If interrupts are desired, set enable bit, RC1IE.
- 4. If 9-bit reception is desired, set bit, RX9.
- 5. Enable the reception by setting bit, CREN.
- Flag bit, RC1IF, will be set when reception is complete and an interrupt will be generated if enable bit, RC1IE, was set.
- 7. Read the RCSTA1 register to get the 9th bit (if enabled) and determine if any error occurred during reception.
- 8. Read the 8-bit received data by reading the RCREG1 register.
- 9. If any error occurred, clear the error by clearing enable bit CREN.
- 10. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

17.3.3 SETTING UP 9-BIT MODE WITH ADDRESS DETECT

This mode would typically be used in RS-485 systems. To set up an Asynchronous Reception with Address Detect Enable:

- 1. Initialize the SPBRGH1:SPBRG1 registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
- 2. Enable the asynchronous serial port by clearing the SYNC bit and setting the SPEN bit.
- If interrupts are required, set the RCEN bit and select the desired priority level with the RC1IP bit.
- 4. Set the RX9 bit to enable 9-bit reception.
- 5. Set the ADDEN bit to enable address detect.
- 6. Enable reception by setting the CREN bit.
- The RC1IF bit will be set when reception is complete. The interrupt will be Acknowledged if the RC1IE and GIE bits are set.
- 8. Read the RCSTA1 register to determine if any error occurred during reception, as well as read bit 9 of data (if applicable).
- 9. Read RCREG1 to determine if the device is being addressed.
- 10. If any error occurred, clear the CREN bit.
- 11. If the device has been addressed, clear the ADDEN bit to allow all received data into the receive buffer and interrupt the CPU.

FIGURE 17-6: EUSART RECEIVE BLOCK DIAGRAM

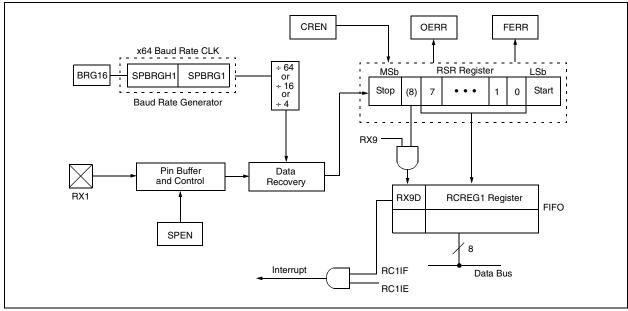


FIGURE 17-7: ASYNCHRONOUS RECEPTION

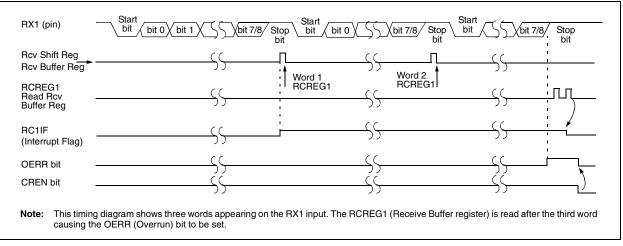


TABLE 17-6: REGISTERS ASSOCIATED WITH ASYNCHRONOUS RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page	
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51	
PIR1	—	ADIF	RC1IF	TX1IF	SSPIF	—	TMR2IF	TMR1IF	54	
PIE1	—	ADIE	RC1IE	TX1IE	SSPIE	—	TMR2IE	TMR1IE	54	
IPR1	—	ADIP	RC1IP	TX1IP	SSPIP	—	TMR2IP	TMR1IP	54	
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	53	
RCREG1	EUSART F	leceive Regi	ster						53	
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	53	
BAUDCON1	ABDOVF	RCMT	_	SCKP	BRG16	—	WUE	ABDEN	55	
SPBRGH1	EUSART E	SART Baud Rate Generator Register High Byte								
SPBRG1	EUSART E	aud Rate G	enerator Re	gister Low	Byte				53	

Legend: — = unimplemented locations read as '0'. Shaded cells are not used for asynchronous reception.

17.3.4 AUTO-WAKE-UP ON SYNC BREAK CHARACTER

During Sleep mode, all clocks to the EUSART are suspended. Because of this, the Baud Rate Generator is inactive and a proper byte reception cannot be performed. The auto-wake-up feature allows the controller to wake-up, due to activity on the RX1/DT1 line, while the EUSART is operating in Asynchronous mode.

The auto-wake-up feature is enabled by setting the WUE bit (BAUDCON<1>). Once set, the typical receive sequence on RX1/DT1 is disabled and the EUSART remains in an Idle state, monitoring for a wake-up event independent of the CPU mode. A wake-up event consists of a high-to-low transition on the RX1/DT1 line. (This coincides with the start of a Sync Break or a Wake-up Signal character for the LIN protocol.)

Following a wake-up event, the module generates an RC1IF interrupt. The interrupt is generated synchronously to the Q clocks in normal operating modes (Figure 17-8) and asynchronously, if the device is in Sleep mode (Figure 17-9). The interrupt condition is cleared by reading the RCREG1 register.

The WUE bit is automatically cleared once a low-to-high transition is observed on the RX1 line following the wake-up event. At this point, the EUSART module is in Idle mode and returns to normal operation. This signals to the user that the Sync Break event is over.

17.3.4.1 Special Considerations Using Auto-Wake-up

Since auto-wake-up functions by sensing rising edge transitions on RX1/DT1, information with any state changes before the Stop bit may signal a false

end-of-character and cause data or framing errors. Therefore, to work properly, the initial character in the transmission must be all '0's. This can be 00h (8 bytes) for standard RS-232 devices, or 000h (12 bits) for LIN bus.

Oscillator start-up time must also be considered, especially in applications using oscillators with longer start-up intervals (i.e., XT or HS mode). The Sync Break (or Wake-up Signal) character must be of sufficient length and be followed by a sufficient interval to allow enough time for the selected oscillator to start and provide proper initialization of the EUSART.

17.3.4.2 Special Considerations Using the WUE Bit

The timing of WUE and RC1IF events may cause some confusion when it comes to determining the validity of received data. As noted, setting the WUE bit places the EUSART in an Idle mode. The wake-up event causes a receive interrupt by setting the RC1IF bit. The WUE bit is cleared after this when a rising edge is seen on RX1/DT1. The interrupt condition is then cleared by reading the RCREG1 register. Ordinarily, the data in RCREG1 will be dummy data and should be discarded.

The fact that the WUE bit has been cleared (or is still set) and the RC1IF flag is set should not be used as an indicator of the integrity of the data in RCREG1. Users should consider implementing a parallel method in firmware to verify received data integrity.

To assure that no actual data is lost, check the RCMT bit to verify that a receive operation is not in process. If a receive operation is not occurring, the WUE bit may then be set just prior to entering the Sleep mode.

FIGURE 17-8: AUTO-WAKE-UP BIT (WUE) TIMINGS DURING NORMAL OPERATION

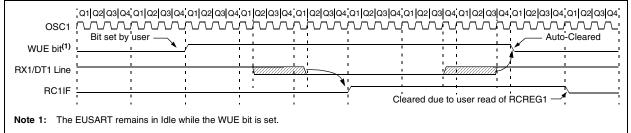
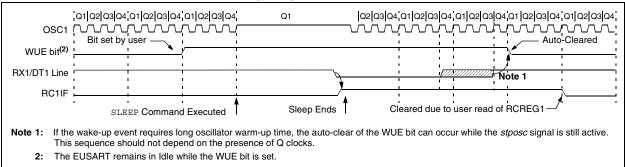


FIGURE 17-9: AUTO-WAKE-UP BIT (WUE) TIMINGS DURING SLEEP



17.3.5 BREAK CHARACTER SEQUENCE

The Enhanced USART module has the capability of sending the special Break character sequences that are required by the LIN bus standard. The Break character transmit consists of a Start bit, followed by twelve '0' bits and a Stop bit. The Frame Break character is sent whenever the SENDB and TXEN bits (TXSTA<3> and TXSTA<5>) are set while the Transmit Shift register is loaded with data. Note that the value of data written to TXREG1 will be ignored and all '0's will be transmitted.

The SENDB bit is automatically reset by hardware after the corresponding Stop bit is sent. This allows the user to preload the transmit FIFO with the next transmit byte following the Break character (typically, the Sync character in the LIN specification).

Note that the data value written to the TXREG1 for the Break character is ignored. The write simply serves the purpose of initiating the proper sequence.

The TRMT bit indicates when the transmit operation is active or Idle, just as it does during normal transmission. See Figure 17-10 for the timing of the Break character sequence.

17.3.5.1 Break and Sync Transmit Sequence

The following sequence will send a message frame header made up of a Break, followed by an Auto-Baud Sync byte. This sequence is typical of a LIN bus master.

- 1. Configure the EUSART for the desired mode.
- 2. Set the TXEN and SENDB bits to set up the Break character.
- 3. Load the TXREG1 with a dummy character to initiate transmission (the value is ignored).
- 4. Write '55h' to TXREG1 to load the Sync character into the transmit FIFO buffer.
- 5. After the Break has been sent, the SENDB bit is reset by hardware. The Sync character now transmits in the preconfigured mode.

When the TXREG1 becomes empty, as indicated by the TX1IF, the next data byte can be written to TXREG1.

17.3.6 RECEIVING A BREAK CHARACTER

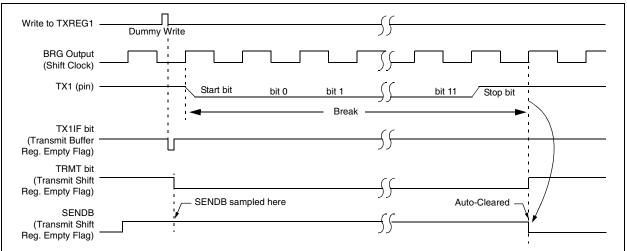
The Enhanced USART module can receive a Break character in two ways.

The first method forces configuration of the baud rate at a frequency of 9/13 the typical speed. This allows for the Stop bit transition to be at the correct sampling location (13 bits for Break versus Start bit and 8 data bits for typical data).

The second method uses the auto-wake-up feature described in **Section 17.3.4** "**Auto-Wake-up On Sync Break Character**". By enabling this feature, the EUSART will sample the next two transitions on RX1/DT1, cause an RC1IF interrupt and receive the next data byte followed by another interrupt.

Note that following a Break character, the user will typically want to enable the Auto-Baud Rate Detect feature. For both methods, the user can set the ABD bit once the TX1IF interrupt is observed.

FIGURE 17-10: SEND BREAK CHARACTER SEQUENCE



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17.4 EUSART Synchronous Master Mode

The Synchronous Master mode is entered by setting the CSRC bit (TXSTA<7>). In this mode, the data is transmitted in a half-duplex manner (i.e., transmission and reception do not occur at the same time). When transmitting data, the reception is inhibited and vice versa. Synchronous mode is entered by setting bit SYNC (TXSTA<4>). In addition, enable bit, SPEN (RCSTA1<7>), is set in order to configure the TX1 and RX1 pins to CK1 (clock) and DT1 (data) lines, respectively.

The Master mode indicates that the processor transmits the master clock on the CK1 line. Clock polarity is selected with the SCKP bit (BAUDCON<4>). Setting SCKP sets the Idle state on CK1 as high, while clearing the bit sets the Idle state as low. This option is provided to support Microwire devices with this module.

17.4.1 EUSART SYNCHRONOUS MASTER TRANSMISSION

The EUSART transmitter block diagram is shown in Figure 17-3. The heart of the transmitter is the Transmit (Serial) Shift register (TSR). The Shift register obtains its data from the Read/Write Transmit Buffer register, TXREG1. The TXREG1 register is loaded with data in software. The TSR register is not loaded until the last bit has been transmitted from the previous load. As soon as the last bit is transmitted, the TSR is loaded with new data from the TXREG1 (if available).

Once the TXREG1 register transfers the data to the TSR register (occurs in one TCYCLE), the TXREG1 is empty and the TX1IF flag bit (PIR1<4>) is set. The interrupt can be enabled or disabled by setting or clearing the interrupt enable bit, TX1IE (PIE1<4>). TX1IF is set regardless of the state of enable bit TX1IE; it cannot be cleared in software. It will reset only when new data is loaded into the TXREG1 register.

While flag bit TX1IF indicates the status of the TXREG1 register, another bit, TRMT (TXSTA<1>), shows the status of the TSR register. TRMT is a read-only bit which is set when the TSR is empty. No interrupt logic is tied to this bit so the user has to poll this bit in order to determine if the TSR register is empty. The TSR is not mapped in data memory so it is not available to the user.

To set up a Synchronous Master Transmission:

- 1. Initialize the SPBRGH1:SPBRG1 registers for the appropriate baud rate. Set or clear the BRG16 bit, as required, to achieve the desired baud rate.
- 2. Enable the synchronous master serial port by setting bits SYNC, SPEN and CSRC.
- 3. If interrupts are desired, set enable bit, TX1IE.
- 4. If 9-bit transmission is desired, set bit, TX9.
- 5. Enable the transmission by setting bit, TXEN.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit TX9D.
- 7. Start transmission by loading data to the TXREG1 register.
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

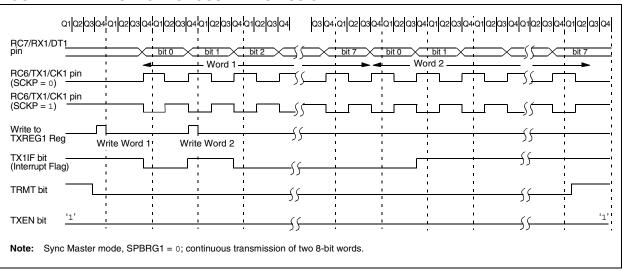


FIGURE 17-11: SYNCHRONOUS TRANSMISSION

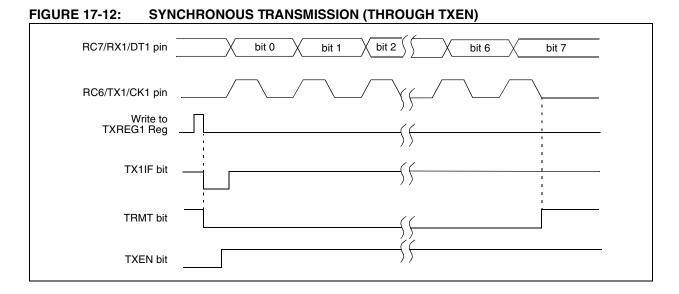


TABLE 17-7: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51
PIR1	—	ADIF	RC1IF	TX1IF	SSPIF	—	TMR2IF	TMR1IF	54
PIE1	—	ADIE	RC1IE	TX1IE	SSPIE	—	TMR2IE	TMR1IE	54
IPR1	—	ADIP	RC1IP	TX1IP	SSPIP	—	TMR2IP	TMR1IP	54
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	53
TXREG1	EUSART T	ransmit Reg	ister						53
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	53
BAUDCON1	ABDOVF	RCMT	_	SCKP	BRG16	—	WUE	ABDEN	55
SPBRGH1	EUSART E	Baud Rate G	enerator Re	gister High	Byte				55
SPBRG1	EUSART E	Baud Rate G	enerator Re	gister Low	Byte				53
LATG	U2OD	U10D	_	LATG4	LATG3	LATG2	LATG1	LATG0	54

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous master transmission.

EUSART SYNCHRONOUS 17.4.2 MASTER RECEPTION

Once Synchronous mode is selected, reception is enabled by setting either the Single Receive Enable bit, SREN (RCSTA1<5>), or the Continuous Receive Enable bit, CREN (RCSTA1<4>). Data is sampled on the RX1 pin on the falling edge of the clock.

If enable bit SREN is set, only a single word is received. If enable bit CREN is set, the reception is continuous until CREN is cleared. If both bits are set, then CREN takes precedence.

To set up a Synchronous Master Reception:

- Initialize the SPBRGH1:SPBRG1 registers for the 1. appropriate baud rate. Set or clear the BRG16 bit, as required, to achieve the desired baud rate.
- 2. Enable the synchronous master serial port by setting bits, SYNC, SPEN and CSRC.

- 3. Ensure bits, CREN and SREN, are clear.
- 4. If interrupts are desired, set enable bit, RC1IE.
- If 9-bit reception is desired, set bit, RX9. 5.
- 6. If a single reception is required, set bit, SREN. For continuous reception, set bit, CREN.
- Interrupt flag bit, RC1IF, will be set when recep-7. tion is complete and an interrupt will be generated if the enable bit, RC1IE, was set.
- Read the RCSTA1 register to get the 9th bit (if 8 enabled) and determine if any error occurred during reception.
- 9. Read the 8-bit received data by reading the RCREG1 register.
- 10. If any error occurred, clear the error by clearing bit, CREN.
- 11. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

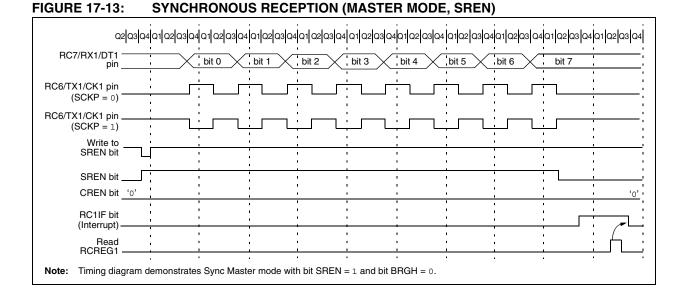


TABLE 17-8: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page	
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51	
PIR1	—	ADIF	RC1IF	TX1IF	SSPIF	—	TMR2IF	TMR1IF	54	
PIE1	—	ADIE	RC1IE	TX1IE	SSPIE	—	TMR2IE	TMR1IE	54	
IPR1	—	ADIP	RC1IP	TX1IP	SSPIP	—	TMR2IP	TMR1IP	54	
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	53	
RCREG1	EUSART R	leceive Regi	ster						53	
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	53	
BAUDCON1	ABDOVF	RCMT		SCKP	BRG16	—	WUE	ABDEN	55	
SPBRGH1	EUSART B	USART Baud Rate Generator Register High Byte								
SPBRG1	PBRG1 EUSART Baud Rate Generator Register Low Byte									
Legend: -	– unimplen	nented read	as '0' Shar	ded cells are	not used fo	or synchron	us master i	recention		

egena: = unimplemented, read as 10°. Snaded cells are not used for synchronous master reception.

17.5 EUSART Synchronous Slave Mode

Synchronous Slave mode is entered by clearing bit CSRC (TXSTA<7>). This mode differs from the Synchronous Master mode in that the shift clock is supplied externally at the CK1 pin (instead of being supplied internally in Master mode). This allows the device to transfer or receive data while in any low-power mode.

17.5.1 EUSART SYNCHRONOUS SLAVE TRANSMIT

The operation of the Synchronous Master and Slave modes are identical except in the case of the Sleep mode.

If two words are written to the TXREG1 and then the SLEEP instruction is executed, the following will occur:

- a) The first word will immediately transfer to the TSR register and transmit.
- b) The second word will remain in the TXREG1 register.
- c) Flag bit, TX1IF, will not be set.
- d) When the first word has been shifted out of TSR, the TXREG1 register will transfer the second word to the TSR and flag bit, TX1IF, will now be set.
- e) If enable bit, TX1IE, is set, the interrupt will wake the chip from Sleep. If the global interrupt is enabled, the program will branch to the interrupt vector.

To set up a Synchronous Slave Transmission:

- 1. Enable the synchronous slave serial port by setting bits SYNC and SPEN and clearing bit CSRC.
- 2. Clear bits CREN and SREN.
- 3. If interrupts are desired, set enable bit, TX1IE.
- 4. If 9-bit transmission is desired, set bit, TX9.
- 5. Enable the transmission by setting enable bit, TXEN.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit TX9D.
- 7. Start transmission by loading data to the TXREG1 register.
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51
PIR1	_	ADIF	RC1IF	TX1IF	SSPIF	_	TMR2IF	TMR1IF	54
PIE1	_	ADIE	RC1IE	TX1IE	SSPIE	_	TMR2IE	TMR1IE	54
IPR1	_	ADIP	RC1IP	TX1IP	SSPIP	—	TMR2IP	TMR1IP	54
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	53
TXREG1	EUSART Transmit Register								
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	53
BAUDCON1	ABDOVF	RCMT		SCKP	BRG16	—	WUE	ABDEN	55
SPBRGH1	EUSART Baud Rate Generator Register High Byte								
SPBRG1	EUSART Baud Rate Generator Register Low Byte								
LATG	U2OD	U10D		LATG4	LATG3	LATG2	LATG1	LATG0	54

TABLE 17-9: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE TRANSMISSION

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous slave transmission.

17.5.2 EUSART SYNCHRONOUS SLAVE RECEPTION

The operation of the Synchronous Master and Slave modes is identical except in the case of Sleep or any Idle mode, and bit SREN, which is a "don't care" in Slave mode.

If receive is enabled by setting the CREN bit prior to entering Sleep or any Idle mode, then a word may be received while in this low-power mode. Once the word is received, the RSR register will transfer the data to the RCREG1 register; if the RC1IE enable bit is set, the interrupt generated will wake the chip from the low-power mode. If the global interrupt is enabled, the program will branch to the interrupt vector. To set up a Synchronous Slave Reception:

- 1. Enable the synchronous master serial port by setting bits SYNC and SPEN and clearing bit CSRC.
- 2. If interrupts are desired, set enable bit, RC1IE.
- 3. If 9-bit reception is desired, set bit, RX9.
- 4. To enable reception, set enable bit, CREN.
- 5. Flag bit, RC1IF, will be set when reception is complete. An interrupt will be generated if enable bit, RC1IE, was set.
- 6. Read the RCSTA1 register to get the 9th bit (if enabled) and determine if any error occurred during reception.
- 7. Read the 8-bit received data by reading the RCREG1 register.
- 8. If any error occurred, clear the error by clearing bit, CREN.
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51
PIR1	—	ADIF	RC1IF	TX1IF	SSPIF	—	TMR2IF	TMR1IF	54
PIE1	—	ADIE	RC1IE	TX1IE	SSPIE	—	TMR2IE	TMR1IE	54
IPR1	—	ADIP	RC1IP	TX1IP	SSPIP	—	TMR2IP	TMR1IP	54
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	53
RCREG1	EUSART Receive Register								
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	53
BAUDCON1	ABDOVF	RCMT		SCKP	BRG16	—	WUE	ABDEN	55
SPBRGH1	EUSART Baud Rate Generator Register High Byte								55
SPBRG1	EUSART Baud Rate Generator Register Low Byte								53

TABLE 17-10: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE RECEPTION

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous slave reception.

18.0 ADDRESSABLE UNIVERSAL SYNCHRONOUS ASYNCHRONOUS RECEIVER TRANSMITTER (AUSART)

The Addressable Universal Synchronous Asynchronous Receiver Transmitter (AUSART) module is very similar in function to the Enhanced USART module, discussed in the previous chapter. It is provided as an additional channel for serial communication with external devices, for those situations that do not require auto-baud detection or LIN bus support.

The AUSART can be configured in the following modes:

- Asynchronous (full-duplex)
- Synchronous Master (half-duplex)
- Synchronous Slave (half-duplex)

The pins of the AUSART module are multiplexed with the functions of PORTG (RG1/TX2/CK2 and RG2/RX2/DT2/VLCAP1, respectively). In order to configure these pins as an AUSART:

- bit SPEN (RCSTA2<7>) must be set (= 1)
- bit TRISG<2> must be set (= 1)
- bit TRISG<1> must be cleared (= 0) for Asynchronous and Synchronous Master modes
- bit TRISG<1> must be set (= 1) for Synchronous Slave mode

Note: The AUSART control will automatically reconfigure the pin from input to output as needed.

The driver for the TX2 output pin can also be optionally configured as an open-drain output. This feature allows the voltage level on the pin to be pulled to a higher level through an external pull-up resistor, and allows the output to communicate with external circuits without the need for additional level shifters.

The open-drain output option is controlled by the U2OD bit (LATG<7>). Setting the bit configures the pin for open-drain operation.

18.1 Control Registers

The operation of the Addressable USART module is controlled through two registers, TXSTA2 and RXSTA2. These are detailed in Register 18-1 and Register 18-2, respectively.

REGISTER R/W-0	R/W-0	R/W-0	R/W-0	STATUS AN U-0	R/W-0	R-1	R/W-0	
CSRC	TX9	TXEN ⁽¹⁾	SYNC		BRGH	TRMT	TX9D	
bit 7	17.9	IXEN	3110		Bridin	TTUNT	bit	
Legend:								
R = Readab	le bit	W = Writable	bit	U = Unimpler	mented bit, rea	ad as '0'		
-n = Value a	t POR	'1' = Bit is set		'0' = Bit is cle		x = Bit is unknown		
bit 7	CSRC: Clock	Source Select	bit					
	<u>Asynchronou</u> Don't care.	<u>s mode:</u>						
		<u>mode:</u> ode (clock gene de (clock from						
bit 6	TX9: 9-Bit Transmit Enable bit 1 = Selects 9-bit transmission 0 = Selects 8-bit transmission							
bit 5	TXEN: Trans	mit Enable bit ⁽¹)					
	1 = Transmit 0 = Transmit							
bit 4	SYNC: AUSA	RT Mode Sele	ct bit					
	1 = Synchror 0 = Asynchro							
bit 3	Unimplemented: Read as '0'							
bit 2	-	Baud Rate Sele	ect bit					
	Asynchronous 1 = High spe							
	0 = Low spece							
	Synchronous mode:							
	Unused in thi							
bit 1		mit Shift Regist	er Status bit					
	1 = TSR emp 0 = TSR full	oty						
bit 0	TX9D: 9th bit	of Transmit Da	ta					
	Can be addre	ess/data bit or a	parity bit.					
Note 1: S	REN/CREN ove		. .					

TRANSMIT STATUS AND CONTROL DECISTER TVOTAO.

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-0	R-0	R-x
SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D
bit 7							bit (
Legend:							
R = Readab		W = Writable I	Dit	-	nented bit, read		
-n = Value a	at POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	iown
oit 7	SPEN: Seria	I Port Enable bit					
	1 = Serial p	ort enabled (con	figures RX2/D	T2 and TX2/CK	2 pins as seria	al port pins)	
	•	ort disabled (held					
bit 6		eceive Enable b	it				
		9-bit reception 8-bit reception					
oit 5		e Receive Enab	le bit				
	Asynchronou	<u>us mode</u> :					
	Don't care.						
		s mode – Master single receive	<u>.</u>				
		single receive					
		eared after recep	tion is comple	ete.			
	<u>Synchronous</u> Don't care.	<u>s mode – Slave:</u>					
oit 4	CREN: Cont	inuous Receive	Enable bit				
	Asynchronou 1 = Enables 0 = Disables	receiver					
	Synchronous						
	1 = Enables	continuous rece continuous rec		le bit CREN is o	cleared (CREN	overrides SRE	EN)
oit 3	ADDEN: Add	dress Detect Ena	able bit				
		us mode 9-bit (R			de the vessive l		
		address detecti address detect					
		us mode 9-bit (R	-				.,
oit 2	FERR: Fram	ing Error bit					
	1 = Framing 0 = No fram	ı error (can be uı ing error	odated by rea	ding RCREG2 r	egister and red	ceiving next val	id byte)
pit 1	OERR: Over	rrun Error bit					
	1 = Overrun 0 = No over	error (can be cl run error	eared by clea	ring bit CREN)			
oit O	RX9D: 9th b	it of Received Da	ata				
	This can be	address/data bit	or a parity hit	and must be as	loulated by us	or firmworo	

18.2 AUSART Baud Rate Generator (BRG)

The BRG is a dedicated, 8-bit generator that supports both the Asynchronous and Synchronous modes of the AUSART.

The SPBRG2 register controls the period of a free-running timer. In Asynchronous mode, bit BRGH (TXSTA<2>) also controls the baud rate. In Synchronous mode, BRGH is ignored. Table 18-1 shows the formula for computation of the baud rate for different AUSART modes, which only apply in Master mode (internally generated clock).

Given the desired baud rate and Fosc, the nearest integer value for the SPBRG2 register can be calculated using the formulas in Table 18-1. From this, the error in baud rate can be determined. An example calculation is shown in Example 18-1. Typical baud rates and error values for the various Asynchronous modes are shown in Table 18-2. It may be advantageous to use the high baud rate (BRGH = 1) to reduce the baud rate error, or achieve a slow baud rate for a fast oscillator frequency.

Writing a new value to the SPBRG2 register causes the BRG timer to be reset (or cleared). This ensures the BRG does not wait for a timer overflow before outputting the new baud rate.

18.2.1 OPERATION IN POWER-MANAGED MODES

The device clock is used to generate the desired baud rate. When one of the power-managed modes is entered, the new clock source may be operating at a different frequency. This may require an adjustment to the value in the SPBRG2 register.

18.2.2 SAMPLING

The data on the RX2 pin is sampled three times by a majority detect circuit to determine if a high or a low level is present at the RX2 pin.

TABLE 18-1:BAUD RATE FORMULAS	
-------------------------------	--

Configur	ation Bits	BRG/AUSART Mode	Baud Rate Formula
SYNC	BRGH	Bha/AUSAn'i Mode	Bauu nale Formula
0	0	Asynchronous	Fosc/[64 (n + 1)]
0	1	Asynchronous	Fosc/[16 (n + 1)]
1	x	Synchronous	Fosc/[4 (n + 1)]

Legend: x = Don't care, n = Value of SPBRG2 register

EXAMPLE 18-1: CALCULATING BAUD RATE ERROR

For a device with FOSO	C of	16 MHz, desired baud rate of 9600, Asynchronous mode, BRGH = 0:
Desired Baud Rate	=	Fosc/(64 ([SPBRG2] + 1))
Solving for SPBRG2:		
Х	=	((FOSC/Desired Baud Rate)/64) - 1
	=	((16000000/9600)/64) - 1
	=	[25.042] = 25
Calculated Baud Rate	=	16000000/(64 (25 + 1))
	=	9615
Error	=	(Calculated Baud Rate - Desired Baud Rate)/Desired Baud Rate
	=	(9615 - 9600)/9600 = 0.16%

TABLE 18-2: REGISTERS ASSOCIATED WITH THE BAUD RATE GENERATOR

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page	
TXSTA2	CSRC	TX9	TXEN	SYNC	_	BRGH	TRMT	TX9D	56	
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	56	
SPBRG2	BRG2 AUSART Baud Rate Generator Register									

Legend: Shaded cells are not used by the BRG.

						BRG	H = 0					
	Fosc	= 40.000) MHz	Fosc = 20.000 MHz			Fosc = 10.000 MHz			Fosc = 8.000 MHz		
BAUD RATE (K)	Actual Rate (K)	% Error	SPBRG value (decimal)									
0.3	—				_	_	_		_		_	_
1.2	—	—	—	1.221	1.73	255	1.202	0.16	129	1.201	-0.16	103
2.4	2.441	1.73	255	2.404	0.16	129	2.404	0.16	64	2.403	-0.16	51
9.6	9.615	0.16	64	9.766	1.73	31	9.766	1.73	15	9.615	-0.16	12
19.2	19.531	1.73	31	19.531	1.73	15	19.531	1.73	7	—	—	_
57.6	56.818	-1.36	10	62.500	8.51	4	52.083	-9.58	2	—	_	_
115.2	125.000	8.51	4	104.167	-9.58	2	78.125	-32.18	1	—	—	—

TABLE 18-3: BAUD RATES FOR ASYNCHRONOUS MODES

		BRGH = 0											
	Fos	c = 4.000	MHz	Fos	c = 2.000	MHz	Fosc = 1.000 MHz						
BAUD RATE (K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)				
0.3	0.300	0.16	207	0.300	-0.16	103	0.300	-0.16	51				
1.2	1.202	0.16	51	1.201	-0.16	25	1.201	-0.16	12				
2.4	2.404	0.16	25	2.403	-0.16	12	—	_	—				
9.6	8.929	-6.99	6	—	_	—	_	_	—				
19.2	20.833	8.51	2	—	—	_	—	_	—				
57.6	62.500	8.51	0	—	_	_	—	_	—				
115.2	62.500	-45.75	0	_	—	—	_	—	—				

		BRGH = 1												
BAUD RATE	Fosc	= 40.000) MHz	Fosc = 20.000 MHz			Fosc	= 10.000) MHz	Fos	Fosc = 8.000 MHz			
(К)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)		
0.3	_	_	_	_	_	_	_	_	_		_	_		
1.2	—	—	—	—	—	—	—	—	—	—	—	—		
2.4	-	_	_	—	_	—	2.441	1.73	255	2.403	-0.16	207		
9.6	9.766	1.73	255	9.615	0.16	129	9.615	0.16	64	9.615	-0.16	51		
19.2	19.231	0.16	129	19.231	0.16	64	19.531	1.73	31	19.230	-0.16	25		
57.6	58.140	0.94	42	56.818	-1.36	21	56.818	-1.36	10	55.555	3.55	8		
115.2	113.636	-1.36	21	113.636	-1.36	10	125.000	8.51	4	—	_	_		

		BRGH = 1											
BAUD RATE	Fosc	= 4.000	MHz	Fos	c = 2.000	MHz	Fos	Fosc = 1.000 MHz					
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)				
0.3	_	_	_	_	_	_	0.300	-0.16	207				
1.2	1.202	0.16	207	1.201	-0.16	103	1.201	-0.16	51				
2.4	2.404	0.16	103	2.403	-0.16	51	2.403	-0.16	25				
9.6	9.615	0.16	25	9.615	-0.16	12	—	—	—				
19.2	19.231	0.16	12	_	—	—	—		—				
57.6	62.500	8.51	3	—	_	_	_	_	_				
115.2	125.000	8.51	1	_	—	—	—		—				

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18.3 AUSART Asynchronous Mode

The Asynchronous mode of operation is selected by clearing the SYNC bit (TXSTA2<4>). In this mode, the AUSART uses standard Non-Return-to-Zero (NRZ) format (one Start bit, eight or nine data bits and one Stop bit). The most common data format is 8 bits. An on-chip, dedicated, 8-bit Baud Rate Generator can be used to derive standard baud rate frequencies from the oscillator.

The AUSART transmits and receives the LSb first. The AUSART's transmitter and receiver are functionally independent but use the same data format and baud rate. The Baud Rate Generator produces a clock, either x16 or x64 of the bit shift rate, depending on the BRGH bit (TXSTA2<2>). Parity is not supported by the hardware but can be implemented in software and stored as the 9th data bit.

When operating in Asynchronous mode, the AUSART module consists of the following important elements:

- Baud Rate Generator
- Sampling Circuit
- Asynchronous Transmitter
- Asynchronous Receiver

18.3.1 AUSART ASYNCHRONOUS TRANSMITTER

The AUSART transmitter block diagram is shown in Figure 18-1. The heart of the transmitter is the Transmit (Serial) Shift register (TSR). The Shift register obtains its data from the Read/Write Transmit Buffer register, TXREG2. The TXREG2 register is loaded with data in software. The TSR register is not loaded until the Stop bit has been transmitted from the previous load. As soon as the Stop bit is transmitted, the TSR is loaded with new data from the TXREG2 register (if available). Once the TXREG2 register transfers the data to the TSR register (occurs in one TCY), the TXREG2 register is empty and the TX2IF flag bit (PIR3<4>) is set. This interrupt can be enabled or disabled by setting or clearing the interrupt enable bit, TX2IE (PIE3<4>). TX2IF will be set regardless of the state of TX2IE; it cannot be cleared in software. TX2IF is also not cleared immediately upon loading TXREG2, but becomes valid in the second instruction cycle following the load instruction. Polling TX2IF immediately following a load of TXREG2 will return invalid results.

While TX2IF indicates the status of the TXREG2 register, another bit, TRMT (TXSTA2<1>), shows the status of the TSR register. TRMT is a read-only bit which is set when the TSR register is empty. No interrupt logic is tied to this bit so the user has to poll this bit in order to determine if the TSR register is empty.

Note 1:	The TSR register is not mapped in data memory so it is not available to the user.
	memory so it is not available to the user.
2:	Flag bit, TX2IF, is set when enable bit, TXEN, is set.

To set up an Asynchronous Transmission:

- 1. Initialize the SPBRG2 register for the appropriate baud rate. Set or clear the BRGH bit, as required, to achieve the desired baud rate.
- 2. Enable the asynchronous serial port by clearing bit SYNC and setting bit SPEN.
- 3. If interrupts are desired, set enable bit, TX2IE.
- 4. If 9-bit transmission is desired, set transmit bit, TX9. Can be used as address/data bit.
- 5. Enable the transmission by setting bit, TXEN, which will also set bit, TX2IF.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit, TX9D.
- 7. Load data to the TXREG2 register (starts transmission).
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

Data Bus TX2IF TXREG2 Register TX2IE 8 MSb Sb Pin Buffer (8) 0 . . . and Control TSR Register TX2 pin Interrupt Baud Rate CLK TXFN SPEN TRMT SPBRG2 TXC Baud Rate Generator TX9D

FIGURE 18-1: AUSART TRANSMIT BLOCK DIAGRAM

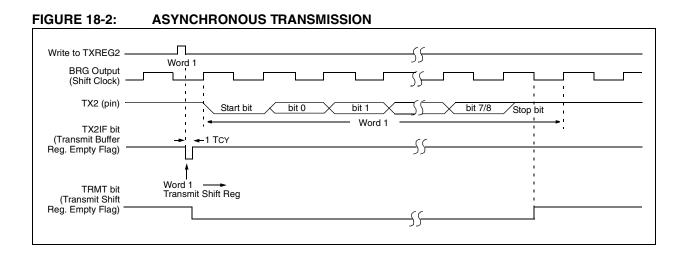


FIGURE 18-3: ASYNCHRONOUS TRANSMISSION (BACK TO BACK)

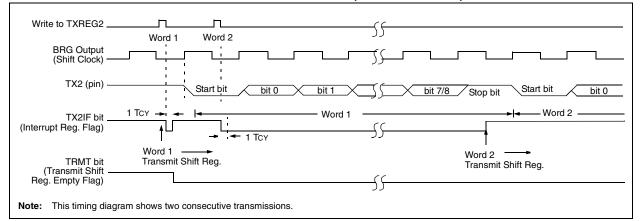


TABLE 18-4: REGISTERS ASSOCIATED WITH ASYNCHRONOUS TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page	
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51	
PIR3	—	LCDIF	RC2IF	TX2IF	—	CCP2IF	CCP1IF	_	54	
PIE3	—	LCDIE	RC2IE	TX2IE		CCP2IE	CCP1IE	_	54	
IPR3	—	LCDIP	RC2IP	TX2IP	_	CCP2IP	CCP1IP	_	54	
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	56	
TXREG2	AUSART T	ransmit Reg	ister						56	
TXSTA2	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	56	
SPBRG2	AUSART Baud Rate Generator Register									
LATG	U2OD	U10D	—	LATG4	LATG3	LATG2	LATG1	LATG0	54	
Laward										

Legend: — = unimplemented locations read as '0'. Shaded cells are not used for asynchronous transmission.

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18.3.2 AUSART ASYNCHRONOUS RECEIVER

The receiver block diagram is shown in Figure 18-4. The data is received on the RX2 pin and drives the data recovery block. The data recovery block is actually a high-speed shifter operating at x16 times the baud rate, whereas the main receive serial shifter operates at the bit rate or at Fosc. This mode would typically be used in RS-232 systems.

To set up an Asynchronous Reception:

- 1. Initialize the SPBRG2 register for the appropriate baud rate. Set or clear the BRGH bit, as required, to achieve the desired baud rate.
- 2. Enable the asynchronous serial port by clearing bit, SYNC, and setting bit, SPEN.
- 3. If interrupts are desired, set enable bit, RC2IE.
- 4. If 9-bit reception is desired, set bit, RX9.
- 5. Enable the reception by setting bit, CREN.
- Flag bit, RC2IF, will be set when reception is complete and an interrupt will be generated if enable bit, RC2IE, was set.
- 7. Read the RCSTA2 register to get the 9th bit (if enabled) and determine if any error occurred during reception.
- 8. Read the 8-bit received data by reading the RCREG2 register.
- 9. If any error occurred, clear the error by clearing enable bit, CREN.
- 10. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

18.3.3 SETTING UP 9-BIT MODE WITH ADDRESS DETECT

This mode would typically be used in RS-485 systems. To set up an Asynchronous Reception with Address Detect Enable:

- 1. Initialize the SPBRG2 register for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
- 2. Enable the asynchronous serial port by clearing the SYNC bit and setting the SPEN bit.
- If interrupts are required, set the RCEN bit and select the desired priority level with the RC2IP bit.
- 4. Set the RX9 bit to enable 9-bit reception.
- 5. Set the ADDEN bit to enable address detect.
- 6. Enable reception by setting the CREN bit.
- 7. The RC2IF bit will be set when reception is complete. The interrupt will be Acknowledged if the RC2IE and GIE bits are set.
- 8. Read the RCSTA2 register to determine if any error occurred during reception, as well as read bit 9 of data (if applicable).
- 9. Read RCREG2 to determine if the device is being addressed.
- 10. If any error occurred, clear the CREN bit.
- 11. If the device has been addressed, clear the ADDEN bit to allow all received data into the receive buffer and interrupt the CPU.

FIGURE 18-4: AUSART RECEIVE BLOCK DIAGRAM

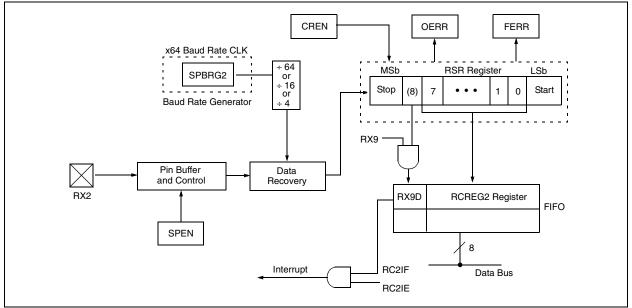


FIGURE 18-5: ASYNCHRONOUS RECEPTION

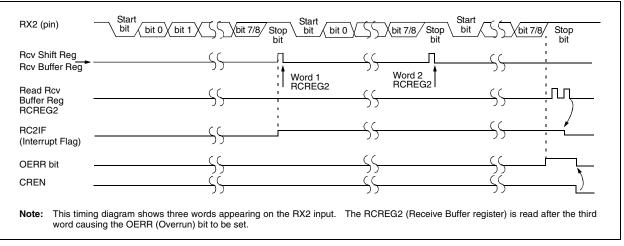


TABLE 18-5: REGISTERS ASSOCIATED WITH ASYNCHRONOUS RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51
PIR3	—	LCDIF	RC2IF	TX2IF	—	CCP2IF	CCP1IF	_	54
PIE3	—	LCDIE	RC2IE	TX2IE	—	CCP2IE	CCP1IE	_	54
IPR3	—	LCDIP	RC2IP	TX2IP	—	CCP2IP	CCP1IP	—	54
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	56
RCREG2	AUSART F	leceive Regi	ster						56
TXSTA2	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	56
SPBRG2	AUSART B	Baud Rate G	enerator Re	gister					56

Legend: — = unimplemented locations read as '0'. Shaded cells are not used for asynchronous reception.

18.4 AUSART Synchronous Master Mode

The Synchronous Master mode is entered by setting the CSRC bit (TXSTA2<7>). In this mode, the data is transmitted in a half-duplex manner (i.e., transmission and reception do not occur at the same time). When transmitting data, the reception is inhibited and vice versa. Synchronous mode is entered by setting bit, SYNC (TXSTA2<4>). In addition, enable bit, SPEN (RCSTA2<7>), is set in order to configure the TX2 and RX2 pins to CK2 (clock) and DT2 (data) lines, respectively.

The Master mode indicates that the processor transmits the master clock on the CK2 line.

18.4.1 AUSART SYNCHRONOUS MASTER TRANSMISSION

The AUSART transmitter block diagram is shown in Figure 18-1. The heart of the transmitter is the Transmit (Serial) Shift register (TSR). The Shift register obtains its data from the Read/Write Transmit Buffer register, TXREG2. The TXREG2 register is loaded with data in software. The TSR register is not loaded until the last bit has been transmitted from the previous load. As soon as the last bit is transmitted, the TSR is loaded with new data from the TXREG2 (if available). Once the TXREG2 register transfers the data to the TSR register (occurs in one TCYCLE), the TXREG2 is empty and the TX2IF flag bit (PIR3<4>) is set. The interrupt can be enabled or disabled by setting or clearing the interrupt enable bit, TX2IE (PIE3<4>). TX2IF is set regardless of the state of enable bit, TX2IE; it cannot be cleared in software. It will reset only when new data is loaded into the TXREG2 register.

While flag bit, TX2IF, indicates the status of the TXREG2 register, another bit, TRMT (TXSTA2<1>), shows the status of the TSR register. TRMT is a read-only bit which is set when the TSR is empty. No interrupt logic is tied to this bit so the user has to poll this bit in order to determine if the TSR register is empty. The TSR is not mapped in data memory so it is not available to the user.

To set up a Synchronous Master Transmission:

- 1. Initialize the SPBRG2 register for the appropriate baud rate.
- 2. Enable the synchronous master serial port by setting bits, SYNC, SPEN and CSRC.
- 3. If interrupts are desired, set enable bit, TX2IE.
- 4. If 9-bit transmission is desired, set bit, TX9.
- 5. Enable the transmission by setting bit, TXEN.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit TX9D.
- 7. Start transmission by loading data to the TXREG2 register.
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

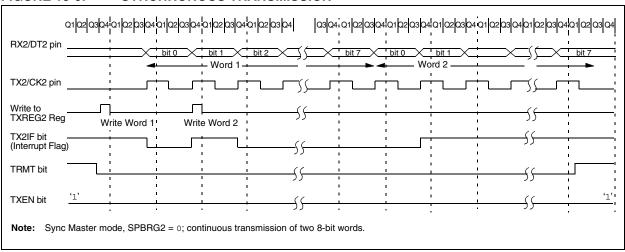


FIGURE 18-6: SYNCHRONOUS TRANSMISSION

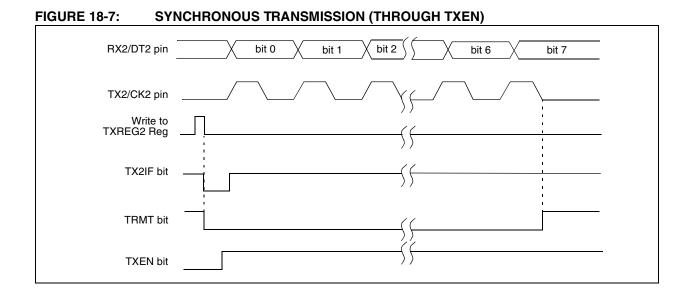


TABLE 18-6: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51
PIR3	—	LCDIF	RC2IF	TX2IF	—	CCP2IF	CCP1IF	_	54
PIE3	—	LCDIE	RC2IE	TX2IE	—	CCP2IE	CCP1IE	_	54
IPR3	—	LCDIP	RC2IP	TX2IP	—	CCP2IP	CCP1IP	_	54
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	56
TXREG2	REG2 AUSART Transmit Register								56
TXSTA2	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	56
SPBRG2 AUSART Baud Rate Generator Register									56
LATG	U2OD	U10D	_	LATG4	LATG3	LATG2	LATG1	LATG0	54

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous master transmission.

18.4.2 AUSART SYNCHRONOUS MASTER RECEPTION

Once Synchronous mode is selected, reception is enabled by setting either the Single Receive Enable bit, SREN (RCSTA2<5>), or the Continuous Receive Enable bit, CREN (RCSTA2<4>). Data is sampled on the RX2 pin on the falling edge of the clock.

If enable bit SREN is set, only a single word is received. If enable bit CREN is set, the reception is continuous until CREN is cleared. If both bits are set, then CREN takes precedence.

To set up a Synchronous Master Reception:

- 1. Initialize the SPBRG2 register for the appropriate baud rate.
- 2. Enable the synchronous master serial port by setting bits SYNC, SPEN and CSRC.
- 3. Ensure bits CREN and SREN are clear.

- 4. If interrupts are desired, set enable bit, RC2IE.
- 5. If 9-bit reception is desired, set bit, RX9.
- 6. If a single reception is required, set bit, SREN. For continuous reception, set bit, CREN.
- 7. Interrupt flag bit, RC2IF, will be set when reception is complete and an interrupt will be generated if the enable bit, RC2IE, was set.
- 8. Read the RCSTA2 register to get the 9th bit (if enabled) and determine if any error occurred during reception.
- 9. Read the 8-bit received data by reading the RCREG2 register.
- 10. If any error occurred, clear the error by clearing bit, CREN.
- 11. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

FIGURE 18-8: SYNCHRONOUS RECEPTION (MASTER MODE, SREN)

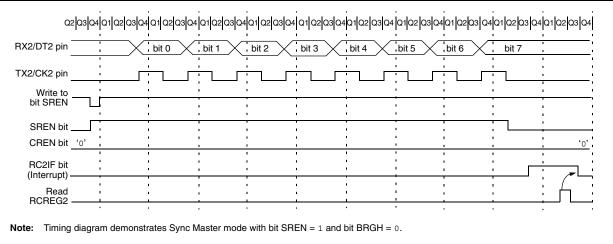


TABLE 18-7: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51
PIR3	—	LCDIF	RC2IF	TX2IF	—	CCP2IF	CCP1IF	—	54
PIE3	—	LCDIE	RC2IE	TX2IE	—	CCP2IE	CCP1IE	_	54
IPR3	—	LCDIP	RC2IP	TX2IP	—	CCP2IP	CCP1IP	_	54
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	56
RCREG2 AUSART Receive Register								56	
TXSTA2	CSRC	TX9	TXEN	SYNC	_	BRGH	TRMT	TX9D	56
SPBRG2 AUSART Baud Rate Generator Register									56

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous master reception.

18.5 AUSART Synchronous Slave Mode

Synchronous Slave mode is entered by clearing bit, CSRC (TXSTA2<7>). This mode differs from the Synchronous Master mode in that the shift clock is supplied externally at the CK2 pin (instead of being supplied internally in Master mode). This allows the device to transfer or receive data while in any low-power mode.

18.5.1 AUSART SYNCHRONOUS SLAVE TRANSMIT

The operation of the Synchronous Master and Slave modes are identical except in the case of the Sleep mode.

If two words are written to the TXREG2 and then the SLEEP instruction is executed, the following will occur:

- a) The first word will immediately transfer to the TSR register and transmit.
- b) The second word will remain in TXREG2 register.
- c) Flag bit, TX2IF, will not be set.
- d) When the first word has been shifted out of TSR, the TXREG2 register will transfer the second word to the TSR and flag bit, TX2IF, will now be set.
- e) If enable bit, TX2IE, is set, the interrupt will wake the chip from Sleep. If the global interrupt is enabled, the program will branch to the interrupt vector.

To set up a Synchronous Slave Transmission:

- 1. Enable the synchronous slave serial port by setting bits, SYNC and SPEN, and clearing bit, CSRC.
- 2. Clear bits, CREN and SREN.
- 3. If interrupts are desired, set enable bit, TX2IE.
- 4. If 9-bit transmission is desired, set bit, TX9.
- 5. Enable the transmission by setting enable bit, TXEN.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit TX9D.
- 7. Start transmission by loading data to the TXREG2 register.
- 8. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51
PIR3	—	LCDIF	RC2IF	TX2IF	—	CCP2IF	CCP1IF	—	54
PIE3		LCDIE	RC2IE	TX2IE	—	CCP2IE	CCP1IE	_	54
IPR3	—	LCDIP	RC2IP	TX2IP	—	CCP2IP	CCP1IP	—	54
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	56
TXREG2	AUSART T	ransmit Regi	ister						56
TXSTA2	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	56
SPBRG2 AUSART Baud Rate Generator Register									56
LATG	U2OD	U10D	_	LATG4	LATG3	LATG2	LATG1	LATG0	54

TABLE 18-8: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE TRANSMISSION

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous slave transmission.

18.5.2 AUSART SYNCHRONOUS SLAVE RECEPTION

The operation of the Synchronous Master and Slave modes is identical except in the case of Sleep or any Idle mode, and bit SREN, which is a "don't care" in Slave mode.

If receive is enabled by setting the CREN bit prior to entering Sleep, or any Idle mode, then a word may be received while in this low-power mode. Once the word is received, the RSR register will transfer the data to the RCREG2 register; if the RC2IE enable bit is set, the interrupt generated will wake the chip from low-power mode. If the global interrupt is enabled, the program will branch to the interrupt vector. To set up a Synchronous Slave Reception:

- 1. Enable the synchronous master serial port by setting bits, SYNC and SPEN, and clearing bit, CSRC.
- 2. If interrupts are desired, set enable bit, RC2IE.
- 3. If 9-bit reception is desired, set bit, RX9.
- 4. To enable reception, set enable bit, CREN.
- 5. Flag bit, RC2IF, will be set when reception is complete. An interrupt will be generated if enable bit, RC2IE, was set.
- 6. Read the RCSTA2 register to get the 9th bit (if enabled) and determine if any error occurred during reception.
- 7. Read the 8-bit received data by reading the RCREG2 register.
- 8. If any error occurred, clear the error by clearing bit, CREN.
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

TABLE 18-9 :	REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51
PIR3	—	LCDIF	RC2IF	TX2IF	—	CCP2IF	CCP1IF	_	54
PIE3		LCDIE	RC2IE	TX2IE	—	CCP2IE	CCP1IE	_	54
IPR3	—	LCDIP	RC2IP	TX2IP	—	CCP2IP	CCP1IP	_	54
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	56
RCREG2	G2 AUSART Receive Register								56
TXSTA2	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	56
SPBRG2 AUSART Baud Rate Generator Register								56	

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous slave reception.

19.0 10-BIT ANALOG-TO-DIGITAL CONVERTER (A/D) MODULE

The Analog-to-Digital (A/D) converter module has 12 inputs for all PIC18F85J90 family devices. This module allows conversion of an analog input signal to a corresponding 10-bit digital number.

The module has five registers:

- A/D Result High Register (ADRESH)
- A/D Result Low Register (ADRESL)
- A/D Control Register 0 (ADCON0)
- A/D Control Register 1 (ADCON1)
- A/D Control Register 2 (ADCON2)

—

The ADCON0 register, shown in Register 19-1, controls the operation of the A/D module. The ADCON1 register, shown in Register 19-2, configures the functions of the port pins. The ADCON2 register, shown in Register 19-3, configures the A/D clock source, programmed acquisition time and justification.

REGISTER 19-1: ADCON0: A/D CONTROL REGISTER 0

R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
ADCAL	—	CHS3	CHS2	CHS1	CHS0	GO/DONE	ADON
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

1 = Calibration is performed on next A/D conversion

- 0 = Normal A/D converter operation (no calibration is performed)
- bit 6 Unimplemented: Read as '0'

	0000 = Channel 00 (AN0)
	0001 = Channel 01 (AN1)
	0010 = Channel 02 (AN2)
	0011 = Channel 03 (AN3)
	0100 = Channel 04 (AN4)
	0101 = Channel 05 (AN5)
	0110 = Channel 06 (AN6)
	0111 = Channel 07 (AN7)
	1000 = Channel 08 (AN8)
	()
	1001 = Channel 09 (AN9)
	1010 = Channel 10 (AN10)
	1011 = Channel 11 (AN11)
	11xx = Unused
bit 1	GO/DONE: A/D Conversion Status bit
	When $ADON = 1$:
	1 = A/D conversion in progress
	0 = A/D Idle
bit 0	ADON: A/D On bit
	1 = A/D converter module is enabled
	0 = A/D converter module is disabled

REGISTER 19-2: ADCON1: A/D CONTROL REGISTER 1

U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	—	VCFG1	VCFG0	PCFG3	PCFG2	PCFG1	PCFG0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6	Unimplemented: Read as '0'
	ommplemented. Head as 0

bit 5	VCFG1: Voltage Reference Configuration bit (VREF- source) 1 = VREF- (AN2) 0 = AVss
bit 4	VCFG0: Voltage Reference Configuration bit (VREF+ source) 1 = VREF+ (AN3) 0 = AVDD

bit 3-0 **PCFG3:PCFG0:** A/D Port Configuration Control bits:

PCFG3: PCFG0	AN11	AN10	AN9	AN8	AN7	AN6	AN5	AN4	AN3	AN2	AN1	AN0
0000	Α	А	А	А	А	А	А	А	А	А	А	Α
0001	Α	А	А	А	А	А	А	А	А	А	А	Α
0010	Α	А	А	А	А	А	А	А	А	А	А	Α
0011	Α	А	А	А	А	А	А	А	А	А	А	Α
0100	D	А	А	А	А	А	А	А	А	А	А	Α
0101	D	D	А	А	А	А	А	А	А	А	А	Α
0110	D	D	D	А	А	А	А	А	А	А	А	Α
0111	D	D	D	D	А	А	А	А	А	А	А	Α
1000	D	D	D	D	D	А	А	А	А	А	А	Α
1001	D	D	D	D	D	D	А	А	А	А	А	Α
1010	D	D	D	D	D	D	D	А	А	А	А	Α
1011	D	D	D	D	D	D	D	D	А	А	А	Α
1100	D	D	D	D	D	D	D	D	D	А	А	Α
1101	D	D	D	D	D	D	D	D	D	D	А	А
1110	D	D	D	D	D	D	D	D	D	D	D	А
1111	D	D	D	D	D	D	D	D	D	D	D	D

A = Analog input

D = Digital I/O

	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0				
ADFM	—	ACQT2	ACQT1	ACQT0	ADCS2	ADCS1	ADCS0				
bit 7							bit C				
Lonordi											
Legend: R = Readab	ala hit	W = Writable	hit	II – Unimplom	aantad hit raa	d oo 'O'					
n = neauar -n = Value a		'1' = Bit is set	DIL	'0' = Bit is clea	nented bit, read	x = Bit is unkr	0000				
		I = DILIS SEL			areu		IOWII				
oit 7	ADFM: A/D R	lesult Format S	elect bit								
	1 = Right justi 0 = Left justifi										
oit 6	-	ted: Read as ')'								
oit 5-3	ACQT2:ACQT0: A/D Acquisition Time Select bits										
	$111 = 20 \text{ TAD}$ $110 = 16 \text{ TAD}$ $101 = 12 \text{ TAD}$ $100 = 8 \text{ TAD}$ $011 = 6 \text{ TAD}$ $010 = 4 \text{ TAD}$ $001 = 2 \text{ TAD}$ $000 = 0 \text{ TAD}^{(1)}$)									
oit 2-0	ADCS2:ADC	S0: A/D Conve	rsion Clock Se	elect bits							
	110 = Fosc/6 101 = Fosc/1 100 = Fosc/4	6 ock derived fro 2		·							

REGISTER 19-3: ADCON2: A/D CONTROL REGISTER 2

a conversion.

The analog reference voltage is software selectable to either the device's positive and negative supply voltage (AVDD and AVSS), or the voltage level on the RA3/AN3/VREF+ and RA2/AN2/VREF- pins.

The A/D converter has a unique feature of being able to operate while the device is in Sleep mode. To operate in Sleep, the A/D conversion clock must be derived from the A/D's internal RC oscillator.

The output of the sample and hold is the input into the converter, which generates the result via successive approximation.

Each port pin associated with the A/D converter can be configured as an analog input or as a digital I/O. The ADRESH and ADRESL registers contain the result of

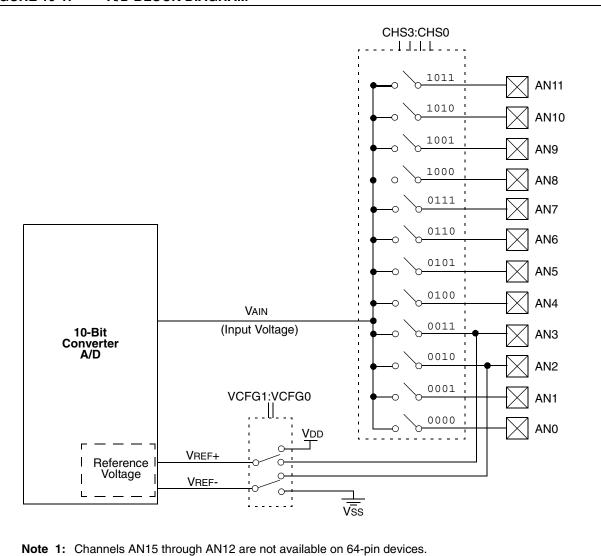


FIGURE 19-1: A/D BLOCK DIAGRAM^(1,2)

the A/D conversion. When the A/D conversion is complete, the result is loaded into the ADRESH:ADRESL register pair, the GO/DONE bit (ADCON0<1>) is cleared and A/D Interrupt Flag bit, ADIF, is set.

A device Reset forces all registers to their Reset state. This forces the A/D module to be turned off and any conversion in progress is aborted. The value in the ADRESH:ADRESL register pair is not modified for a Power-on Reset. These registers will contain unknown data after a Power-on Reset.

The block diagram of the A/D module is shown in Figure 19-1.

Iterational Science Sc

After the A/D module has been configured as desired, the selected channel must be acquired before the conversion is started. The analog input channels must have their corresponding TRIS bits selected as an input. To determine acquisition time, see **Section 19.1** "**A/D Acquisition Requirements**". After this acquisition time has elapsed, the A/D conversion can be started. An acquisition time <u>can be</u> programmed to occur between setting the GO/DONE bit and the actual start of the conversion.

The following steps should be followed to do an A/D conversion:

- 1. Configure the A/D module:
 - Configure analog pins, voltage reference and digital I/O (ADCON1)
 - Select A/D input channel (ADCON0)
 - Select A/D acquisition time (ADCON2)
 - Select A/D conversion clock (ADCON2)
 - Turn on A/D module (ADCON0)
- 2. Configure A/D interrupt (if desired):
 - Clear ADIF bit
 - Set ADIE bit
 - Set GIE bit

- 3. Wait the required acquisition time (if required).
- 4. Start conversion:
 Set GO/DONE bit (ADCON0<1>)
- 5. Wait for A/D conversion to complete, by either:
 Polling for the GO/DONE bit to be cleared OR
 - Waiting for the A/D interrupt
- 6. Read A/D Result registers (ADRESH:ADRESL); clear ADIF bit, if required.
- 7. For next conversion, go to step 1 or step 2, as required. The A/D conversion time per bit is defined as TAD. A minimum wait of 2 TAD is required before next acquisition starts.

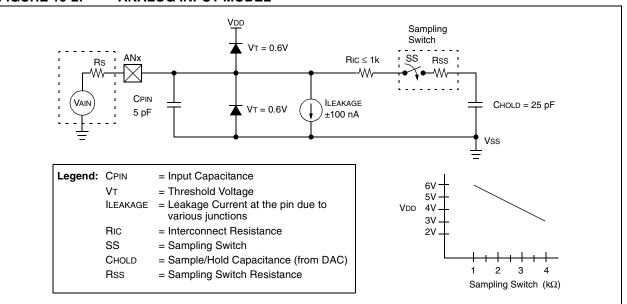


FIGURE 19-2: ANALOG INPUT MODEL

19.1 A/D Acquisition Requirements

For the A/D converter to meet its specified accuracy, the charge holding capacitor (CHOLD) must be allowed to fully charge to the input channel voltage level. The analog input model is shown in Figure 19-2. The source impedance (Rs) and the internal sampling switch (Rss) impedance directly affect the time required to charge the capacitor CHOLD. The sampling switch (Rss) impedance varies over the device voltage (VDD). The source impedance affects the offset voltage at the analog input (due to pin leakage current). The maximum recommended impedance for analog sources is 2.5 k Ω . After the analog input channel is selected (changed), the channel must be sampled for at least the minimum acquisition time before starting a conversion.

Note:	When	the	conversion	is	started,	the
	holding) capa	acitor is disco	nne	ected from	the
	input p	in.				

EQUATION 19-1: ACQUISITION TIME

TACQ = Amplifier Settling Time + Holding Capacitor Charging Time + Temperature Coefficient = TAMP + TC + TCOFF

EQUATION 19-2: A/D MINIMUM CHARGING TIME

VHOLD	=	$(\text{VREF} - (\text{VREF}/2048)) \bullet (1 - e^{(-\text{TC/CHOLD}(\text{Ric} + \text{Rss} + \text{Rs}))})$
or		
TC	=	-(Chold)(Ric + Rss + Rs) ln(1/2048)

EQUATION 19-3: CALCULATING THE MINIMUM REQUIRED ACQUISITION TIME

TACQ	=	TAMP + TC + TCOFF
TAMP	=	0.2 μs
TCOFF	=	(Temp – 25°C)(0.02 μs/°C) (85°C – 25°C)(0.02 μs/°C) 1.2 μs
Tempera	ature c	oefficient is only required for temperatures $> 25^{\circ}$ C. Below 25° C, TCOFF = 0 ms.
ТС	=	-(Chold)(Ric + Rss + Rs) $\ln(1/2048) \ \mu s$ -(25 pF) (1 k Ω + 2 k Ω + 2.5 k Ω) ln(0.0004883) μs 1.05 μs
TACQ	=	$0.2 \ \mu s + 1 \ \mu s + 1.2 \ \mu s$ $2.4 \ \mu s$

To calculate the minimum acquisition time, Equation 19-1 may be used. This equation assumes that 1/2 LSb error is used (1024 steps for the A/D). The 1/2 LSb error is the maximum error allowed for the A/D to meet its specified resolution.

Equation 19-3 shows the calculation of the minimum required acquisition time, TACQ. This calculation is based on the following application system assumptions:

=	25 pF
=	2.5 kΩ
\leq	1/2 LSb
=	$3V \rightarrow Rss = 2 \ k\Omega$
=	85°C (system max.)
	= ≤ =

19.2 Selecting and Configuring Automatic Acquisition Time

The ADCON2 register allows the user to select an acquisition time that occurs each time the GO/DONE bit is set.

When the GO/DONE bit is set, sampling is stopped and a conversion begins. The user is responsible for ensuring the required acquisition time has passed between selecting the desired input channel and setting the GO/DONE bit. This occurs when the ACQT2:ACQT0 bits (ADCON2<5:3>) remain in their Reset state ('000') and is compatible with devices that do not offer programmable acquisition times.

If desired, the ACQT bits can be set to select a programmable acquisition time for the A/D module. When the GO/DONE bit is set, the A/D module continues to sample the input for the selected acquisition time, then automatically begins a conversion. Since the acquisition time is programmed, there may be no need to wait for an acquisition time between selecting a channel and setting the GO/DONE bit.

In either case, when the conversion is completed, the GO/DONE bit is cleared, the ADIF flag is set and the A/D begins sampling the currently selected channel again. If an acquisition time is programmed, there is nothing to indicate if the acquisition time has ended or if the conversion has begun.

19.3 Selecting the A/D Conversion Clock

The A/D conversion time per bit is defined as TAD. The A/D conversion requires 11 TAD per 10-bit conversion. The source of the A/D conversion clock is software selectable.

There are seven possible options for TAD:

- 2 Tosc
- 4 Tosc
- 8 Tosc
- 16 Tosc
- 32 Tosc
- 64 Tosc
- Internal RC Oscillator

For correct A/D conversions, the A/D conversion clock (TAD) must be as short as possible but greater than the minimum TAD (see parameter 130 in Table 25-25 for more information).

Table 19-1 shows the resultant TAD times derived from the device operating frequencies and the A/D clock source selected.

TABLE 19-1: TAD vs. DEVICE OPERATING FREQUENCIES

AD Clock S	AD Clock Source (TAD)					
Operation	ADCS2:ADCS0	Device Frequency				
2 Tosc	000	2.86 MHz				
4 Tosc	100	5.71 MHz				
8 Tosc	001	11.43 MHz				
16 Tosc	101	22.86 MHz				
32 Tosc	010	40.0 MHz				
64 Tosc	110	40.0 MHz				
RC ⁽²⁾	x11	1.00 MHz ⁽¹⁾				

Note 1: The RC source has a typical TAD time of $4 \mu s$.

2: For device frequencies above 1 MHz, the device must be in Sleep mode for the entire conversion or the A/D accuracy may be out of specification.

19.4 Configuring Analog Port Pins

The ADCON1, TRISA, TRISF and TRISH registers control the operation of the A/D port pins. The port pins needed as analog inputs must have their corresponding TRIS bits set (input). If the TRIS bit is cleared (output), the digital output level (VOH or VOL) will be converted.

The A/D operation is independent of the state of the CHS3:CHS0 bits and the TRIS bits.

- Note 1: When reading the PORT register, all pins configured as analog input channels will read as cleared (a low level). Pins configured as digital inputs will convert an analog input. Analog levels on a digitally configured input will be accurately converted.
 - 2: Analog levels on any pin defined as a digital input may cause the digital input buffer to consume current out of the device's specification limits.

19.5 A/D Conversions

Figure 19-3 shows the operation of the A/D converter after the GO/DONE bit has been set and the ACQT2:ACQT0 bits are cleared. A conversion is started after the following instruction to allow entry into Sleep mode before the conversion begins.

Figure 19-4 shows the operation of the A/D converter after the GO/DONE bit has been set, the ACQT2:ACQT0 bits are set to '010' and selecting a 4 TAD acquisition time before the conversion starts.

Clearing the GO/DONE bit during a conversion will abort the current conversion. The A/D Result register pair will NOT be updated with the partially completed A/D conversion sample. This means the ADRESH:ADRESL registers will continue to contain the value of the last completed conversion (or the last value written to the ADRESH:ADRESL registers).

After the A/D conversion is completed or aborted, a 2 TAD wait is required before the next acquisition can be started. After this wait, acquisition on the selected channel is automatically started.

Note:	The GO/DONE bit should NOT be set in
	the same instruction that turns on the A/D.

19.6 Use of the CCP2 Trigger

An A/D conversion can be started by the "Special Event Trigger" of the CCP2 module. This requires that the CCP2M3:CCP2M0 bits (CCP2CON<3:0>) be programmed as '1011' and that the A/D module is enabled (ADON bit is set). When the trigger occurs, the GO/DONE bit will be set, starting the A/D acquisition and conversion and the Timer1 (or Timer3) counter will be reset to zero. Timer1 (or Timer3) is reset to automatically repeat the A/D acquisition period with minimal software overhead (moving ADRESH/ADRESL to the desired location). The appropriate analog input channel must be selected and the minimum acquisition period is either timed by the user, or an appropriate TACQ time is <u>selected</u> before the Special Event Trigger sets the GO/DONE bit (starts a conversion).

If the A/D module is not enabled (ADON is cleared), the Special Event Trigger will be ignored by the A/D module but will still reset the Timer1 (or Timer3) counter.

FIGURE 19-3: A/D CONVERSION TAD CYCLES (ACQT2:ACQT0 = 000, TACQ = 0)

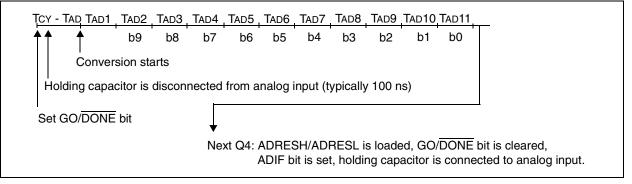
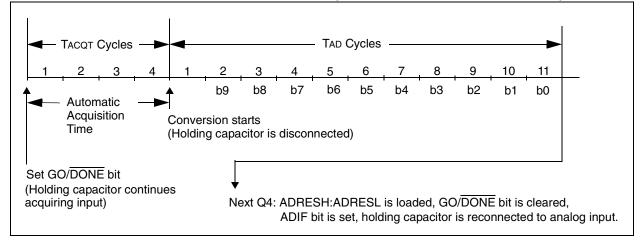


FIGURE 19-4: A/D CONVERSION TAD CYCLES (ACQT2:ACQT0 = 010, TACQ = 4 TAD)



19.7 A/D Converter Calibration

The A/D converter in the PIC18F85J90 family of devices includes a self-calibration feature which compensates for any offset generated within the module. The calibration process is automated and is initiated by setting the ADCAL bit (ADCON0<7>). The next time the GO/DONE bit is set, the module will perform a "dummy" conversion (that is, with reading none of the input channels) and store the resulting value internally to compensate for offset. Thus, subsequent offsets will be compensated.

The calibration process assumes that the device is in a relatively steady-state operating condition. If A/D calibration is used, it should be performed after each device Reset or if there are other major changes in operating conditions.

19.8 Operation in Power-Managed Modes

The selection of the automatic acquisition time and A/D conversion clock is determined in part by the clock source and frequency while in a power-managed mode.

If the A/D is expected to operate while the device is in a power-managed mode, the ACQT2:ACQT0 and ADCS2:ADCS0 bits in ADCON2 should be updated in accordance with the power-managed mode clock that will be used. After the power-managed mode is entered (either of the power-managed Run modes), an A/D acquisition or conversion may be started. Once an acquisition or conversion is started, the device should continue to be clocked by the same power-managed mode clock source until the conversion has been completed. If desired, the device may be placed into the corresponding power-managed Idle mode during the conversion.

If the power-managed mode clock frequency is less than 1 MHz, the A/D RC clock source should be selected.

Operation in the Sleep mode requires the A/D RC clock to be selected. If bits ACQT2:ACQT0 are set to '000' and a conversion is started, the conversion will be delayed one instruction cycle to allow execution of the SLEEP instruction and entry to Sleep mode. The IDLEN and SCS bits in the OSCCON register must have already been cleared prior to starting the conversion.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51
PIR1	—	ADIF	RC1IF	TX1IF	SSPIF	—	TMR2IF	TMR1IF	54
PIE1	_	ADIE	RC1IE	TX1IE	SSPIE	—	TMR2IE	TMR1IE	54
IPR1	—	ADIP	RC1IP	TX1IP	SSPIP	—	TMR2IP	TMR1IP	54
PIR3	—	LCDIF	RC2IF	TX2IF	—	CCP2IF	CCP1IF	_	54
PIE3	_	LCDIE	RC2IE	TX2IE	—	CCP2IE	CCP1IE		54
IPR3	_	LCDIP	RC2IP	TX2IP	—	CCP2IP	CCP1IP		54
ADRESH	A/D Result	t Register Hi	gh Byte						53
ADRESL	A/D Result	t Register Lo	w Byte						53
ADCON0	ADCAL	_	CHS3	CHS3	CHS1	CHS0	GO/DONE	ADON	53
ADCON1	—	—	VCFG1	VCFG0	PCFG3	PCFG2	PCFG1	PCFG0	53
ADCON2	ADFM	_	ACQT2	ACQT1	ACQT0	ADCS2	ADCS1	ADCS0	53
CCP2CON	_	_	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	55
PORTA	RA7 ⁽¹⁾	RA6 ⁽¹⁾	RA5	RA4	RA3	RA2	RA1	RA0	55
TRISA	TRISA7 ⁽¹⁾	TRISA6 ⁽¹⁾	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	54
PORTF	RF7	RF6	RF5	RF4	RF3	RF2	RF1	_	54
TRISF	TRISF5	TRISF4	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1	_	54

TABLE 19-2: SUMMARY OF A/D REGISTERS

Legend: — = unimplemented, read as '0'. Shaded cells are not used for A/D conversion.

Note 1: RA6/RA7 and their associated latch and direction bits are configured as port pins only when the internal oscillator is selected as the default clock source (FOSC2 Configuration bit = 0); otherwise, they are disabled and these bits read as '0'.

NOTES:

20.0 COMPARATOR MODULE

The analog comparator module contains two comparators that can be configured in a variety of ways. The inputs can be selected from the analog inputs multiplexed with pins RF1 through RF6, as well as the on-chip voltage reference (see Section 21.0 "Comparator Voltage Reference Module"). The digital outputs (normal or inverted) are available at the pin level and can also be read through the control register.

The CMCON register (Register 20-1) selects the comparator input and output configuration. Block diagrams of the various comparator configurations are shown in Figure 20-1.

REGISTER 20-1: CMCON: COMPARATOR MODULE CONTROL REGISTER

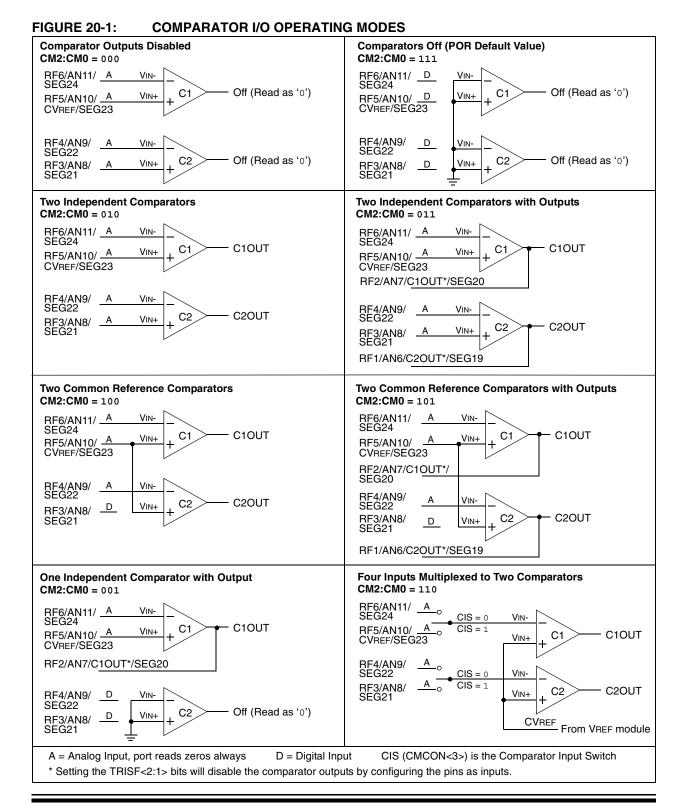
R-0	R-0	R/W-0	R/W-0	R/W-0	R/W-1	R/W-1	R/W-1
C2OUT	C1OUT	C2INV	C1INV	CIS	CM2	CM1	CM0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown
<u>Wher</u> 1 = C 0 = C	JT : Comparator 2 Output bit C2INV = 0: 2 VIN+ > C2 VIN- 2 VIN+ < C2 VIN- C2INV = 1:		
1 = C	2 Vin+ < C2 Vin- 2 Vin+ > C2 Vin-		
<u>Wher</u> 1 = C 0 = C	JT : Comparator 1 Output bit C1INV = 0: 1 VIN+ > C1 VIN- 1 VIN+ < C1 VIN- C1INV = 1:		
$0 = \mathbf{C}$	1 Vin+ < C1 Vin- 1 Vin+ > C1 Vin-		
1 = C	 Comparator 2 Output Inversion bi output inverted output not inverted 	T	
1 = C	 Comparator 1 Output Inversion bi 1 output inverted 1 output not inverted 	t	
When 1 = 0 0 = 0	Comparator Input Switch bit <u>CM2:CM0 = 110:</u> C1 VIN- connects to RF5/AN10/CVRE C2 VIN- connects to RF3/AN8 C1 VIN- connects to RF6/AN11 C2 VIN- connects to RF4/AN9	F	
	CM0 : Comparator Mode bits e 20-1 shows the Comparator mode	s and the CM2:CM0 bit set	ttings.

20.1 Comparator Configuration

There are eight modes of operation for the comparators, shown in Figure 20-1. Bits CM2:CM0 of the CMCON register are used to select these modes. The TRISF register controls the data direction of the comparator pins for each mode. If the Comparator mode is changed, the comparator output level may not be valid for the specified mode change delay shown in **Section 25.0 "Electrical Characteristics"**.

Note:	Comparator interrupts should be disabled
	during a Comparator mode change;
	otherwise, a false interrupt may occur.



20.2 Comparator Operation

A single comparator is shown in Figure 20-2, along with the relationship between the analog input levels and the digital output. When the analog input at VIN+ is less than the analog input VIN-, the output of the comparator is a digital low level. When the analog input at VIN+ is greater than the analog input VIN-, the output of the comparator is a digital high level. The shaded areas of the output of the comparator in Figure 20-2 represent the uncertainty due to input offsets and response time.

20.3 Comparator Reference

Depending on the comparator operating mode, either an external or internal voltage reference may be used. The analog signal present at VIN- is compared to the signal at VIN+ and the digital output of the comparator is adjusted accordingly (Figure 20-2).

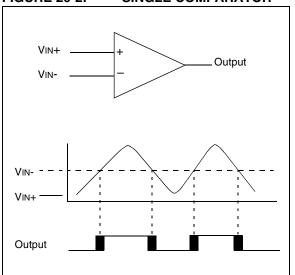


FIGURE 20-2: SINGLE COMPARATOR

20.3.1 EXTERNAL REFERENCE SIGNAL

When external voltage references are used, the comparator module can be configured to have the comparators operate from the same or different reference sources. However, threshold detector applications may require the same reference. The reference signal must be between VSS and VDD and can be applied to either pin of the comparator(s).

20.3.2 INTERNAL REFERENCE SIGNAL

The comparator module also allows the selection of an internally generated voltage reference from the comparator voltage reference module. This module is described in more detail in **Section 21.0 "Comparator Voltage Reference Module"**.

The internal reference is only available in the mode where four inputs are multiplexed to two comparators (CM2:CM0 = 110). In this mode, the internal voltage reference is applied to the VIN+ pin of both comparators.

20.4 Comparator Response Time

Response time is the minimum time, after selecting a new reference voltage or input source, before the comparator output has a valid level. If the internal reference is changed, the maximum delay of the internal voltage reference must be considered when using the comparator outputs. Otherwise, the maximum delay of the comparators should be used (see Section 25.0 "Electrical Characteristics").

20.5 Comparator Outputs

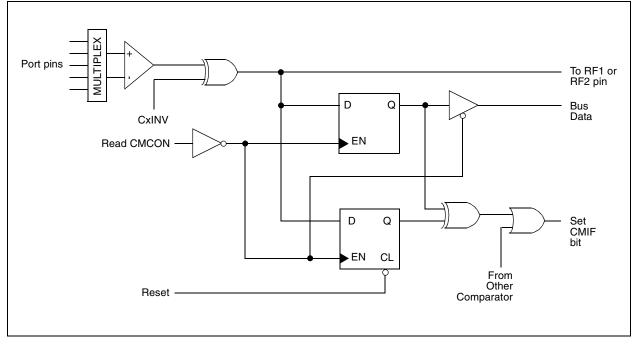
The comparator outputs are read through the CMCON register. These bits are read-only. The comparator outputs may also be directly output to the RF1 and RF2 I/O pins. When enabled, multiplexors in the output path of the RF1 and RF2 pins will switch and the output of each pin will be the unsynchronized output of the comparator. The uncertainty of each of the comparators is related to the input offset voltage and the response time given in the specifications. Figure 20-3 shows the comparator output block diagram.

The TRISF bits will still function as an output enable/ disable for the RF1 and RF2 pins while in this mode.

The polarity of the comparator outputs can be changed using the C2INV and C1INV bits (CMCON<5:4>).

- Note 1: When reading the PORT register, all pins configured as analog inputs will read as '0'. Pins configured as digital inputs will convert an analog input according to the Schmitt Trigger input specification.
 - 2: Analog levels on any pin defined as a digital input may cause the input buffer to consume more current than is specified.

FIGURE 20-3: COMPARATOR OUTPUT BLOCK DIAGRAM



20.6 Comparator Interrupts

The comparator interrupt flag is set whenever there is a change in the output value of either comparator. Software will need to maintain information about the status of the output bits, as read from CMCON<7:6>, to determine the actual change that occurred. The CMIF bit (PIR2<6>) is the Comparator Interrupt Flag. The CMIF bit must be reset by clearing it. Since it is also possible to write a '1' to this register, a simulated interrupt may be initiated.

Both the CMIE bit (PIE2<6>) and the PEIE bit (INTCON<6>) must be set to enable the interrupt. In addition, the GIE bit (INTCON<7>) must also be set. If any of these bits are clear, the interrupt is not enabled, though the CMIF bit will still be set if an interrupt condition occurs.

Note:	If a change in the CMCON register								
	(C1OUT or C2OUT) should occur when a								
	read operation is being executed (start of								
	the Q2 cycle), then the CMIF (PIR2<6>)								
	interrupt flag may not get set.								

The user, in the Interrupt Service Routine, can clear the interrupt in the following manner:

- a) Any read or write of CMCON will end the mismatch condition.
- b) Clear flag bit CMIF.

A mismatch condition will continue to set flag bit CMIF. Reading CMCON will end the mismatch condition and allow flag bit CMIF to be cleared.

20.7 Comparator Operation During Sleep

When a comparator is active and the device is placed in Sleep mode, the comparator remains active and the interrupt is functional, if enabled. This interrupt will wake-up the device from Sleep mode, when enabled. Each operational comparator will consume additional current, as shown in the comparator specifications. To minimize power consumption while in Sleep mode, turn off the comparators (CM2:CM0 = 111) before entering Sleep. If the device wakes up from Sleep, the contents of the CMCON register are not affected.

20.8 Effects of a Reset

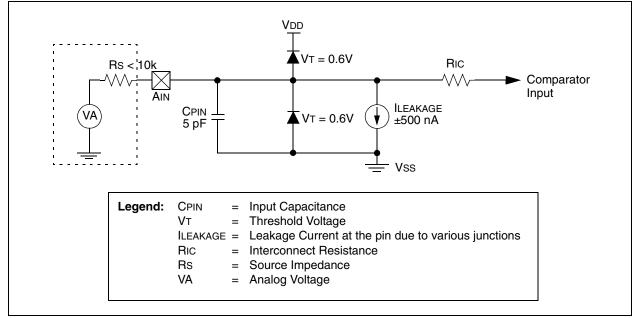
A device Reset forces the CMCON register to its Reset state, causing the comparator modules to be turned off (CM2:CM0 = 111). However, the input pins (RF3 through RF6) are configured as analog inputs by default on device Reset. The I/O configuration for these pins is determined by the setting of the PCFG3:PCFG0 bits (ADCON1<3:0>). Therefore, device current is minimized when analog inputs are present at Reset time.

20.9 Analog Input Connection Considerations

A simplified circuit for an analog input is shown in Figure 20-4. Since the analog pins are connected to a digital output, they have reverse biased diodes to VDD and Vss. The analog input, therefore, must be between Vss and VDD. If the input voltage deviates from this

range by more than 0.6V in either direction, one of the diodes is forward biased and a latch-up condition may occur. A maximum source impedance of 10 k Ω is recommended for the analog sources. Any external component connected to an analog input pin, such as a capacitor or a Zener diode, should have very little leakage current.





Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	51
PIR2	OSCFIF	CMIF	_	_	BCLIF	LVDIF	TMR3IF	_	54
PIE2	OSCFIE	CMIE	_	_	BCLIE	LVDIE	TMR3IE	—	54
IPR2	OSCFIP	CMIP			BCLIP	LVDIP	TMR3IP		54
CMCON	C2OUT	C10UT	C2INV	C1INV	CIS	CM2	CM1	CM0	53
CVRCON	CVREN	CVROE	CVRR	CVRSS	CVR3	CVR2	CVR1	CVR0	53
PORTF	RF7	RF6	RF5	RF4	RF3	RF2	RF1	_	54
LATF	LATF7	LATF6	LATF5	LATF4	LATF3	LATF2	LATF1	_	54
TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1	_	54

Legend: — = unimplemented, read as '0'. Shaded cells are unused by the comparator module.

NOTES:

21.0 COMPARATOR VOLTAGE REFERENCE MODULE

The comparator voltage reference is a 16-tap resistor ladder network that provides a selectable reference voltage. Although its primary purpose is to provide a reference for the analog comparators, it may also be used independently of them.

A block diagram of the module is shown in Figure 21-1. The resistor ladder is segmented to provide two ranges of CVREF values and has a power-down function to conserve power when the reference is not being used. The module's supply reference can be provided from either device VDD/VSS or an external voltage reference.

21.1 Configuring the Comparator Voltage Reference

The comparator voltage reference module is controlled through the CVRCON register (Register 21-1). The comparator voltage reference provides two ranges of output voltage, each with 16 distinct levels. The range to be used is selected by the CVRR bit (CVRCON<5>). The primary difference between the ranges is the size of the steps selected by the CVREF Selection bits (CVR3:CVR0), with one range offering finer resolution. The equations used to calculate the output of the comparator voltage reference are as follows:

<u>If CVRR = 1:</u> CVREF = ((CVR3:CVR0)/24) x (CVRSRC) <u>If CVRR = 0:</u> CVREF = (CVRSRC/4) + ((CVR3:CVR0)/32) x (CVRSRC)

The comparator reference supply voltage can come from either VDD and VSS, or the external VREF+ and VREF- that are multiplexed with RA2 and RA3. The voltage source is selected by the CVRSS bit (CVRCON<4>).

The settling time of the comparator voltage reference must be considered when changing the CVREF output (see Table 25-3 in Section 25.0 "Electrical Characteristics").

REGISTER 21-1: CVRCON: COMPARATOR VOLTAGE REFERENCE CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
CVREN	CVROE ⁽¹⁾	CVRR	CVRSS	CVR3	CVR2	CVR1	CVR0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

CVREN: Comparator Voltage Reference Enable bit
1 = CVREF circuit powered on
0 = CVREF circuit powered down
CVROE: Comparator VREF Output Enable bit ⁽¹⁾
 1 = CVREF voltage level is also output on the RF5/AN10/CVREF/SEG23 pin 0 = CVREF voltage is disconnected from the RF5/AN10/CVREF/SEG23 pin
CVRR: Comparator VREF Range Selection bit
1 = 0 to 0.667 CVRSRC, with CVRSRC/24 step size (low range)
0 = 0.25 CVRSRC to 0.75 CVRSRC, with CVRSRC/32 step size (high range)
CVRSS: Comparator VREF Source Selection bit
1 = Comparator reference source, CVRSRC = (VREF+) - (VREF-)
0 = Comparator reference source, CVRSRC = VDD - VSS
CVR3:CVR0: Comparator VREF Value Selection bits ($0 \le (CVR3:CVR0) \le 15$)
When CVRR = 1:
$CVREF = ((CVR3:CVR0)/24) \bullet (CVRSRC)$
When CVRR = 0:
$CVREF = (CVRSRC/4) + ((CVR3:CVR0)/32) \bullet (CVRSRC)$

Note 1: CVROE overrides the TRISF<5> bit setting.

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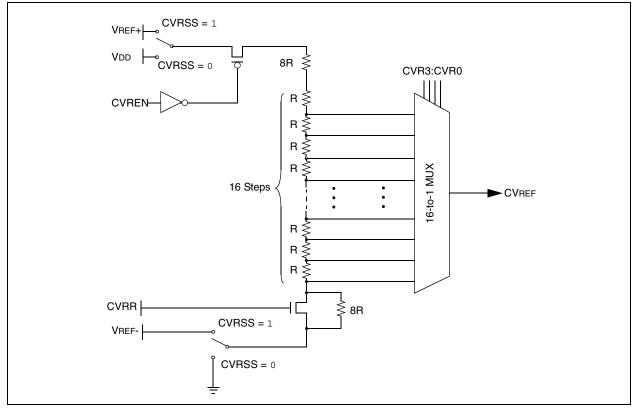


FIGURE 21-1: COMPARATOR VOLTAGE REFERENCE BLOCK DIAGRAM

21.2 Voltage Reference Accuracy/Error

The full range of voltage reference cannot be realized due to the construction of the module. The transistors on the top and bottom of the resistor ladder network (Figure 21-1) keep CVREF from approaching the reference source rails. The voltage reference is derived from the reference source; therefore, the CVREF output changes with fluctuations in that source. The tested absolute accuracy of the voltage reference can be found in **Section 25.0 "Electrical Characteristics"**.

21.3 Operation During Sleep

When the device wakes up from Sleep through an interrupt or a Watchdog Timer time-out, the contents of the CVRCON register are not affected. To minimize current consumption in Sleep mode, the voltage reference should be disabled.

21.4 Effects of a Reset

A device Reset disables the voltage reference by clearing bit, CVREN (CVRCON<7>). This Reset also disconnects the reference from the RA2 pin by clearing bit, CVROE (CVRCON<6>) and selects the high-voltage range by clearing bit, CVRR (CVRCON<5>). The CVR value select bits are also cleared.

21.5 Connection Considerations

The voltage reference module operates independently of the comparator module. The output of the reference generator may be connected to the RF5 pin if the CVROE bit is set. Enabling the voltage reference output onto RA2 when it is configured as a digital input will increase current consumption. Connecting RF5 as a digital output with CVRSS enabled will also increase current consumption.

The RF5 pin can be used as a simple D/A output with limited drive capability. Due to the limited current drive capability, a buffer must be used on the voltage reference output for external connections to VREF. Figure 21-2 shows an example buffering technique.

FIGURE 21-2: COMPARATOR VOLTAGE REFERENCE OUTPUT BUFFER EXAMPLE

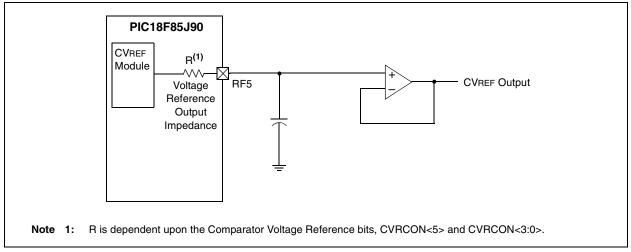


TABLE 21-1: REGISTERS ASSOCIATED WITH COMPARATOR VOLTAGE REFERENCE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
CVRCON	CVREN	CVROE	CVRR	CVRSS	CVR3	CVR2	CVR1	CVR0	53
CMCON	C2OUT	C10UT	C2INV	C1INV	CIS	CM2	CM1	CM0	53
TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1		54

Legend: — = unimplemented, read as '0'. Shaded cells are not used with the comparator voltage reference.

NOTES:

22.0 SPECIAL FEATURES OF THE CPU

PIC18F85J90 family devices include several features intended to maximize reliability and minimize cost through elimination of external components. These are:

- Oscillator Selection
- Resets:
 - Power-on Reset (POR)
 - Power-up Timer (PWRT)
 - Oscillator Start-up Timer (OST)
 - Brown-out Reset (BOR)
- Interrupts
- Watchdog Timer (WDT)
- Fail-Safe Clock Monitor
- Two-Speed Start-up
- Code Protection
- In-Circuit Serial Programming

The oscillator can be configured for the application depending on frequency, power, accuracy and cost. All of the options are discussed in detail in **Section 2.0 "Oscillator Configurations"**.

A complete discussion of device Resets and interrupts is available in previous sections of this data sheet.

In addition to their Power-up and Oscillator Start-up Timers provided for Resets, the PIC18F85J90 family family of devices have a configurable Watchdog Timer which is controlled in software.

The inclusion of an internal RC oscillator also provides the additional benefits of a Fail-Safe Clock Monitor (FSCM) and Two-Speed Start-up. FSCM provides for background monitoring of the peripheral clock and automatic switchover in the event of its failure. Two-Speed Start-up enables code to be executed almost immediately on start-up, while the primary clock source completes its start-up delays.

All of these features are enabled and configured by setting the appropriate Configuration register bits.

22.1 Configuration Bits

The Configuration bits can be programmed (read as '0'), or left unprogrammed (read as '1'), to select various device configurations. These bits are mapped starting at program memory location 300000h. A complete list is shown in Table 22-2. A detailed explanation of the various bit functions is provided in Register 22-1 through Register 22-5.

22.1.1 CONSIDERATIONS FOR CONFIGURING THE PIC18F85J90 FAMILY DEVICES

Devices of the PIC18F85J90 family do not use persistent memory registers to store configuration information. The configuration bytes are implemented as volatile memory which means that configuration data must be programmed each time the device is powered up.

Configuration data is stored in the four words at the top of the on-chip program memory space, known as the Flash Configuration Words. It is stored in program memory in the same order shown in Table 22-2, with CONFIG1L at the lowest address and CONFIG3H at the highest. The data is automatically loaded in the proper Configuration registers during device power-up.

When creating applications for these devices, users should always specifically allocate the location of the Flash Configuration Word for configuration data. This is to make certain that program code is not stored in this address when the code is compiled.

The volatile memory cells used for the Configuration bits always reset to '1' on Power-on Resets. For all other types of Reset events, the previously programmed values are maintained and used without reloading from program memory.

The four Most Significant bits of CONFIG1H, CONFIG2H and CONFIG3H in program memory should also be '1111'. This makes these Configuration Words appear to be NOP instructions in the remote event that their locations are ever executed by accident. Since Configuration bits are not implemented in the corresponding locations, writing '1's to these locations has no effect on device operation.

To prevent inadvertent configuration changes during code execution, all programmable Configuration bits are write-once. After a bit is initially programmed during a power cycle, it cannot be written to again. Changing a device configuration requires that power to the device be cycled.

MAPPING OF THE FLASH

THE CONFIGURATION

CONFIGURATION WORDS TO

TABLE 22-1:

REGISTERS								
Configuration Byte	Code Space Address	Configuration Register Address						
CONFIG1L	XXXF8h	300000h						
CONFIG1H	XXXF9h	300001h						
CONFIG2L	XXXFAh	300002h						
CONFIG2H	XXXFBh	300003h						
CONFIG3L	XXXFCh	300004h						
CONFIG3H	XXXFDh	300005h						
Legende Unimplemented in PIC19E95 100 family								

Legend: Unimplemented in PIC18F85J90 family devices.

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TABLE 22-2: CONFIGURATION BITS AND DEVICE IDs

File	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Default/ Unprogrammed Value ⁽¹⁾
300000h	CONFIG1L	DEBUG	XINST	STVREN	_	_	—	_	WDTEN	1111
300001h	CONFIG1H	(2)	(2)	(2)	(2)	(3)	CP0	_	—	01
300002h	CONFIG2L	IESO	FCMEN	—	—	—	FOSC2	FOSC1	FOSC0	11111
300003h	CONFIG2H	(2)	(2)	(2)	(2)	WDTPS3	WDTPS2	WDTPS1	WDTPS0	1111
300005h	CONFIG3H	(2)	(2)	(2)	(2)	—	_	—	CCP2MX	1
3FFFFEh	DEVID1	DEV2	DEV1	DEV0	REV4	REV3	REV2	REV1	REV0	xxxx xxxx ⁽⁴⁾
3FFFFFh	DEVID2	DEV10	DEV9	DEV8	DEV7	DEV6	DEV5	DEV4	DEV3	0000 10x1 (4)

Legend: x = unknown, - = unimplemented. Shaded cells are unimplemented, read as '0'.

Note 1: Values reflect the unprogrammed state as received from the factory and following Power-on Resets. In all other Reset states, the configuration bytes maintain their previously programmed states.

2: The value of these bits in program memory should always be '1'. This ensures that the location is executed as a NOP if it is accidentally executed.

3: This bit should always be maintained as '0'.

4: See Register 22-6 and Register 22-7 for DEVID values. These registers are read-only and cannot be programmed by the user.

REGISTER 22-1: CONFIG1L: CONFIGURATION REGISTER 1 LOW (BYTE ADDRESS 300000h)

REGISTER	22-1: CON	FIGIL: CONF	IGURATION	REGISTER		ADDRESS	300000n)			
R/WO-1	R/WO-1	R/WO-1	U-0	U-0	U-0	U-0	R/WO-1			
DEBUG	XINST	STVREN	_	_	—	_	WDTEN			
bit 7				·			bit			
Legend: R = Readabl	e bit	WO = Write-O	nce bit	U = Unimpler	nented bit, read	l as '0'				
-n = Value w	nen device is u	nprogrammed		'1' = Bit is set		'0' = Bit is cle	eared			
bit 7	1 = Backgro 0 = Backgro	kground Debug und debugger d und debugger e	isabled; RB6 a nabled; RB6 a	and RB7 config and RB7 are de	•	• •	pins			
bit 6	 XINST: Extended Instruction Set Enable bit 1 = Instruction set extension and Indexed Addressing mode enabled 0 = Instruction set extension and Indexed Addressing mode disabled (Legacy mode) 									
bit 5	STVREN: Stack Overflow/Underflow Reset Enable bit									
		n stack overflow, n stack overflow,								
			•							

- bit 4-1 Unimplemented: Read as '0'
- bit 0 WDTEN: Watchdog Timer Enable bit
 - 1 = WDT enabled
 - 0 = WDT disabled (control is placed on SWDTEN bit)

REGISTER 22-2: CONFIG1H: CONFIGURATION REGISTER 1 HIGH (BYTE ADDRESS 300001h)

U-0	U-0	U-0	U-0	U-0	R/WO-1	U-0	U-0
(1)	(1)	(1)	(1)	(2)	CP0		_
bit 7							bit 0

Legend:			
R = Readable bit	WO = Write-Once bit	U = Unimplemented b	pit, read as '0'
-n = Value when device	is unprogrammed	'1' = Bit is set	'0' = Bit is cleared

bit 7-3 Unimplemented: Read as '0'

bit 2

- CP0: Code Protection bit
 - 1 = Program memory is not code-protected
 - 0 = Program memory is code-protected
- bit 1-0 Unimplemented: Read as '0'
- **Note 1:** The value of these bits in program memory should always be '1'. This ensures that the location is executed as a NOP if it is accidentally executed.
 - 2: This bit should always be maintained as '0'.

REGISTER 22-3: CONFIG2L: CONFIGURATION REGISTER 2 LOW (BYTE ADDRESS 300002h)

	R/WO-1	U-0	U-0	U-0	R/WO-1	R/WO-1	R/WO-1
IESO	FCMEN		_	_	FOSC2	FOSC1	FOSC0
bit 7							bit 0

Legend:			
R = Readable bit	WO = Write-Once bit	U = Unimplemented bit, read as '0'	
-n = Value when device is unprogrammed		'1' = Bit is set	'0' = Bit is cleared

bit 7	IESO: Two-Speed Start-up (Internal/External Oscillator Switchover) Control bit 1 = Two-Speed Start-up enabled 0 = Two-Speed Start-up disabled
bit 6	FCMEN: Fail-Safe Clock Monitor Enable bit
	 1 = Fail-Safe Clock Monitor enabled 0 = Fail-Safe Clock Monitor disabled
bit 5-3	Unimplemented: Read as '0'
bit 2-0	FOSC2:FOSC0: Oscillator Selection bits
	111 = OSC1/OSC2 as primary; EC oscillator with CLKO function and software controlled PLL (EC+PLL)
	110 = OSC1/OSC2 as primary; EC oscillator with CLKO function (EC)
	101 = OSC1/OSC2 as primary; HS oscillator with software controlled PLL (HS+PLL)
	100 = OSC1/OSC2 as primary; HS oscillator (HS)
	011 = INTOSC with CLKO as primary; port function on RA7; EC oscillator with CLKO function and software controlled PLL (EC+PLL)
	010 = INTOSC with CLKO as primary; port function on RA7; EC oscillator with CLKO function
	001 = INTOSC as primary with port function on RA6/RA7; HS oscillator with software controlled PLL (HS+PLL)

000 = INTOSC as primary with port function on RA6/RA7; HS oscillator (HS)

REGISTER 22-4: CONFIG2H: CONFIGURATION REGISTER 2 HIGH (BYTE ADDRESS 300003h)

				-		-
U-0	U-0	U-0	R/WO-1	R/WO-1	R/WO-1	R/WO-1
(1)	(1)	(1)	WDTPS3	WDTPS2	WDTPS1	WDTPS0
						bit (
e bit	WO = Write-O	nce bit	U = Unimplem	nented bit, read	as '0'	
en device is un	programmed		'1' = Bit is set		'0' = Bit is clea	ared
1111 = 1:32,7 1110 = 1:16,3 1101 = 1:8,19 1100 = 1:4,09 1011 = 1:2,04 1010 = 1:1,02 1001 = 1:512	768 384 92 96 48 24	og Timer Post	scale Select bit	5		
1000 = 1:256						
	(1) bit en device is un Unimplement WDTPS3:WD 1111 = 1:32,7 1110 = 1:16,3 1101 = 1:8,19 1001 = 1:4,09 1011 = 1:2,04 1010 = 1:1,02 1001 = 1:512	(1)(1) bit WO = Write-O en device is unprogrammed Unimplemented: Read as '0	(1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1) enderstand (1) (1) enderstand (1) (1) unimplemented: Read as '0' (1) WDTPS3:WDTPS0: Watchdog Timer Posts 1111 = 1:32,768 (1) (1) 1100 = 1:4,096 (1) (1) 1011 = 1:2,048 (1) (1) 1001 = 1:512 (1) (1)	(1) (1) (1) WDTPS3 bit WO = Write-Once bit U = Unimplem en device is unprogrammed '1' = Bit is set Unimplemented: Read as '0' WDTPS3:WDTPS0: Watchdog Timer Postscale Select bits 111 = 1:32,768 1110 = 1:16,384 1101 = 1:8,192 1100 = 1:4,096 1011 = 1:2,048 1010 = 1:1,024 1001 = 1:512	(1)(1)WDTPS3WDTPS2wbitWO = Write-Once bitU = Unimplemented bit, read (1' = Bit is setunimplemented: Read as '0'WDTPS3:WDTPS0: Watchdog Timer Postscale Select bits111 = 1:32,7681110 = 1:16,3841101 = 1:8,1921100 = 1:4,0961011 = 1:2,0481001 = 1:512	(1) (1) (1) WDTPS3 WDTPS2 WDTPS1 bit $WO = Write-Once bit U = Unimplemented bit, read as '0' en device is unprogrammed '1' = Bit is set '0' = Bit is clear Unimplemented: Read as '0' WDTPS3:WDTPS0: Watchdog Timer Postscale Select bits 1111 = 1:32,768 1110 = 1:16,384 1101 = 1:8,192 1100 = 1:4,096 1011 = 1:2,048 1010 = 1:1,024 1001 = 1:512$

REGISTER 22-5:	CONFIG3H: CONFIGURATION REGISTER 3 HIGH (BYTE ADDRESS 300005h)

Note 1: The value of these bits in program memory should always be '1'. This ensures that the location is

U-0	U-0	U-0	U-0	U-0	U-0	U-0	R/WO-1
(1)	(1)	(1)	(1)	—	—	—	CCP2MX
bit 7							bit 0

Legend:			
R = Readable bit	WO = Write-Once bit	U = Unimplemented b	bit, read as '0'
-n = Value when device	is unprogrammed	'1' = Bit is set	'0' = Bit is cleared

bit 7-1 Unimplemented: Read as '0'

0110 = 1:64 0101 = 1:32 0100 = 1:16 0011 = 1:8 0010 = 1:4 0001 = 1:2 0000 = 1:1

bit 0 CCP2MX: CCP2 MUX bit

- 1 = CCP2 is multiplexed with RC1
- 0 = CCP2 is multiplexed with RE7

executed as a NOP if it is accidentally executed.

Note 1: The value of these bits in program memory should always be '1'. This ensures that the location is executed as a NOP if it is accidentally executed.

REGISTER 22-6: DEVID1: DEVICE ID REGISTER 1 FOR PIC18F85J90 FAMILY DEVICES

R	R	R	R	R	R	R	R
DEV2	DEV1	DEV0	REV4	REV3	REV2	REV1	REV0
bit 7							bit 0

Legend:	
R = Read-only bit	

bit 7-5 **DEV2:DEV0:** Device ID bits

111 = PIC18F85J90
101 = PIC18F84J90
100 = PIC18F83J90
011 = PIC18F65J90
001 = PIC18F64J90
000 = PIC18F63J90

bit 4-0 **REV4:REV0:** Revision ID bits These bits are used to indicate the device revision.

REGISTER 22-7: DEVID2: DEVICE ID REGISTER 2 FOR PIC18F85J90 FAMILY DEVICES

R	R	R	R	R	R	R	R
DEV10 ⁽¹⁾	DEV9 ⁽¹⁾	DEV8 ⁽¹⁾	DEV7 ⁽¹⁾	DEV6 ⁽¹⁾	DEV5 ⁽¹⁾	DEV4 ⁽¹⁾	DEV3 ⁽¹⁾
bit 7							bit 0

Legend:			
R = Read-only bit			

bit 7-0 **DEV10:DEV3:** Device ID bits⁽¹⁾

These bits are used with the DEV2:DEV0 bits in the Device ID Register 1 to identify the part number. 0011 1000 = PIC18F6XJ90/8XJ90 devices

Note 1: The values for DEV10:DEV3 may be shared with other device families. The specific device is always identified by using the entire DEV10:DEV0 bit sequence.

Watchdog Timer (WDT) 22.2

For PIC18F85J90 family devices, the WDT is driven by the INTRC oscillator. When the WDT is enabled, the clock source is also enabled. The nominal WDT period is 4 ms and has the same stability as the INTRC oscillator.

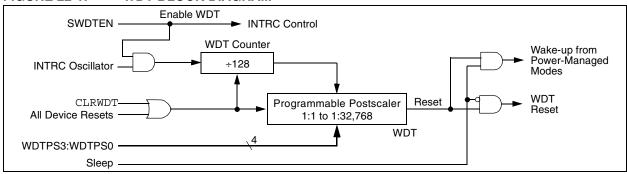
The 4 ms period of the WDT is multiplied by a 16-bit postscaler. Any output of the WDT postscaler is selected by a multiplexor, controlled by the WDTPS bits in Configuration Register 2H. Available periods range from 4 ms to 131.072 seconds (2.18 minutes). The WDT and postscaler are cleared whenever a SLEEP or CLRWDT instruction is executed, or a clock failure (primary or Timer1 oscillator) has occurred.

FIGURE 22-1:	WDT BLOCK DIAGRAM
FIGURE 22-1:	WDT BLOCK DIAGRAM

- Note 1: The CLRWDT and SLEEP instructions clear the WDT and postscaler counts when executed.
 - 2: When a CLRWDT instruction is executed, the postscaler count will be cleared.

22.2.1 CONTROL REGISTER

The WDTCON register (Register 22-8) is a readable and writable register. The SWDTEN bit enables or disables WDT operation. This allows software to override the WDTEN Configuration bit and enable the WDT only if it has been disabled by the Configuration bit.



REGISTER 22-8: WDTCON: WATCHDOG TIMER CONTROL REGISTER

R/W-0	U-0	U-0	U-0	U-0	U-0	U-0	R/W-0
REGSLP ⁽¹⁾	—	—	—	—	—	_	SWDTEN ⁽²⁾
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	REGSLP: Voltage Regulator Low-Power Operation Enable bit ⁽¹⁾ 1 = On-chip regulator enters low-power operation when device enters Sleep mode 0 = On-chip regulator continues to operate normally in Sleep mode
bit 6-1	Unimplemented: Read as '0'
bit 0	SWDTEN: Software Controlled Watchdog Timer Enable bit ⁽²⁾
	 1 = Watchdog Timer is on 0 = Watchdog Timer is off

Note 1: The REGSLP bit is automatically cleared when a Low-Voltage Detect condition occurs.

This bit has no effect if the Configuration bit, WDTEN, is enabled. 2:

TABLE 22-3: SUMMARY OF WATCHDOG TIMER REGISTERS									
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
RCON	IPEN			RI	TO	PD	POR	BOR	52
WDTCON	REGSLP			_				SWDTEN	52

Legend: — = unimplemented, read as '0'. Shaded cells are not used by the Watchdog Timer.

22.3 On-Chip Voltage Regulator

All of the PIC18F85J90 family devices power their core digital logic at a nominal 2.5V. For designs that are required to operate at a higher typical voltage, such as 3.3V, all devices in the PIC18F85J90 family incorporate an on-chip regulator that allows the device to run its core logic from VDD.

The regulator is controlled by the ENVREG pin. Tying VDD to the pin enables the regulator, which in turn, provides power to the core from the other VDD pins. When the regulator is enabled, a low-ESR filter capacitor must be connected to the VDDCORE/VCAP pin (Figure 22-2). This helps to maintain the stability of the regulator. The recommended value for the filter capacitor is provided in Section 25.3 "DC Characteristics: PIC18F84J90 Family (Industrial)".

If ENVREG is tied to Vss, the regulator is disabled. In this case, separate power for the core logic at a nominal 2.5V must be supplied to the device on the VDDCORE/VCAP pin to run the I/O pins at higher voltage levels, typically 3.3V. Alternatively, the VDDCORE/VCAP and VDD pins can be tied together to operate at a lower nominal voltage. Refer to Figure 22-2 for possible configurations.

22.3.1 VOLTAGE REGULATION AND LOW VOLTAGE DETECTION

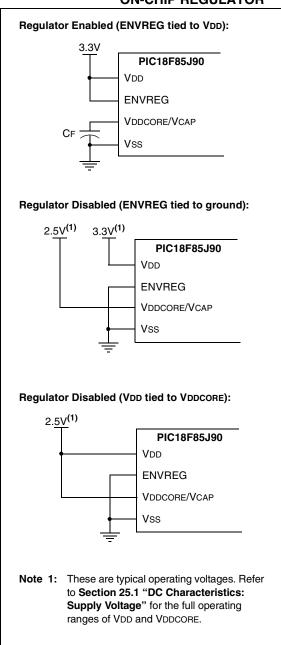
When it is enabled, the on-chip regulator provides a constant voltage of 2.5V nominal to the digital core logic. The regulator can provide this level from a Vdd of about 2.5V, all the way up to the device's Vddmax. It does not have the capability to boost Vdd levels below 2.5V.

In order to prevent "brown-out" conditions, when the voltage drops too low for the regulator, the regulator enters Tracking mode. In Tracking mode, the regulator output follows VDD, with a typical voltage drop of 100 mV.

The on-chip regulator includes a simple Low-Voltage Detect (LVD) circuit. If Vdd drops too low to maintain approximately 2.45V on Vddcore, the circuit sets the Low-Voltage Detect Interrupt Flag, LVDIF (PIR2<2>) and clears the REGSLP (WDTCON<7>) bit, if it was set.

This can be used to generate an interrupt and put the application into a low-power operational mode, or trigger an orderly shutdown. Low-Voltage Detection is only available when the regulator is enabled.

FIGURE 22-2: CONNECTIONS FOR THE ON-CHIP REGULATOR



22.3.2 ON-CHIP REGULATOR AND BOR

When the on-chip regulator is enabled, PIC18F85J90 family devices also have a simple Brown-out Reset capability. If the voltage supplied to the regulator falls to a level that is inadequate to maintain a regulated output for full-speed operation, the regulator Reset circuitry will generate a Brown-out Reset. This event is captured by the BOR flag bit (RCON<0>).

The operation of the BOR is described in more detail in Section 4.4 "Brown-out Reset (BOR)" and Section 4.4.1 "Detecting BOR".

22.3.3 POWER-UP REQUIREMENTS

The on-chip regulator is designed to meet the power-up requirements for the device. If the application does not use the regulator, then strict power-up conditions must be adhered to. While powering up, VDDCORE must never exceed VDD by 0.3 volts.

22.3.4 OPERATION IN SLEEP MODE

When enabled, the on-chip regulator always consumes a small incremental amount of current over IDD. This includes when the device is in Sleep mode, even though the core digital logic does not require power. To provide additional savings in applications where power resources are critical, the regulator can be configured to automatically disable itself whenever the device goes into Sleep mode. This feature is controlled by the REGSLP bit (WDTCON<7>). Setting this bit disables the regulator in Sleep mode, and reduces its current consumption to a minimum.

Substantial Sleep mode power savings can be obtained by setting the REGSLP bit, but device wake-up time will increase in order to ensure the regulator has enough time to stabilize. The REGSLP bit is automatically cleared by hardware when a Low-Voltage Detect condition occurs. The REGSLP bit can be set again in software, which would continue to keep the voltage regulator in Low-Power mode. This, however, is not recommended if any write operations to the Flash will be performed.

22.4 Two-Speed Start-up

The Two-Speed Start-up feature helps to minimize the latency period, from oscillator start-up to code execution, by allowing the microcontroller to use the INTRC oscillator as a clock source until the primary clock source is available. It is enabled by setting the IESO Configuration bit.

Two-Speed Start-up should be enabled only if the primary oscillator mode is HS or HSPLL (Crystal-Based) modes. Since the EC and ECPLL modes do not require an OST start-up delay, Two-Speed Start-up should be disabled.

When enabled, Resets and wake-ups from Sleep mode cause the device to configure itself to run from the internal oscillator block as the clock source, following the time-out of the Power-up Timer after a Power-on Reset is enabled. This allows almost immediate code execution while the primary oscillator starts and the OST is running. Once the OST times out, the device automatically switches to PRI_RUN mode.

In all other power-managed modes, Two-Speed Start-up is not used. The device will be clocked by the currently selected clock source until the primary clock source becomes available. The setting of the IESO bit is ignored.

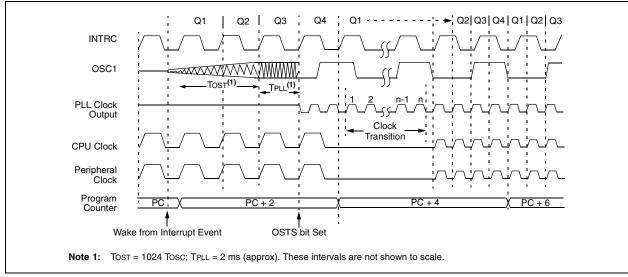


FIGURE 22-3: TIMING TRANSITION FOR TWO-SPEED START-UP (INTRC TO HSPLL)

22.4.1 SPECIAL CONSIDERATIONS FOR USING TWO-SPEED START-UP

While using the INTRC oscillator in Two-Speed Start-up, the device still obeys the normal command sequences for entering power-managed modes, including serial SLEEP instructions (refer to **Section 3.1.4 "Multiple Sleep Commands"**). In practice, this means that user code can change the SCS1:SCS0 bit settings or issue SLEEP instructions before the OST times out. This would allow an application to briefly wake-up, perform routine "housekeeping" tasks and return to Sleep before the device starts to operate from the primary oscillator.

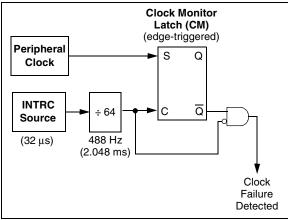
User code can also check if the primary clock source is currently providing the device clocking by checking the status of the OSTS bit (OSCCON<3>). If the bit is set, the primary oscillator is providing the clock. Otherwise, the internal oscillator block is providing the clock during wake-up from Reset or Sleep mode.

22.5 Fail-Safe Clock Monitor

The Fail-Safe Clock Monitor (FSCM) allows the microcontroller to continue operation in the event of an external oscillator failure by automatically switching the device clock to the internal oscillator block. The FSCM function is enabled by setting the FCMEN Configuration bit.

When FSCM is enabled, the INTRC oscillator runs at all times to monitor clocks to peripherals and provides a backup clock in the event of a clock failure. Clock monitoring (shown in Figure 22-4) is accomplished by creating a sample clock signal which is the INTRC output divided by 64. This allows ample time between FSCM sample clocks for a peripheral clock edge to occur. The peripheral device clock and the sample clock are presented as inputs to the Clock Monitor latch (CM). The CM is set on the falling edge of the device clock source but cleared on the rising edge of the sample clock.

FIGURE 22-4: FSCM BLOCK DIAGRAM



Clock failure is tested for on the falling edge of the sample clock. If a sample clock falling edge occurs while CM is still set, a clock failure has been detected (Figure 22-5). This causes the following:

- the FSCM generates an oscillator fail interrupt by setting bit, OSCFIF (PIR2<7>);
- the device clock source is switched to the internal oscillator block (OSCCON is not updated to show the current clock source – this is the fail-safe condition); and
- the WDT is reset.

During switchover, the postscaler frequency from the internal oscillator block may not be sufficiently stable for timing sensitive applications. In these cases, it may be desirable to select another clock configuration and enter an alternate power-managed mode. This can be done to attempt a partial recovery or execute a controlled shutdown. See Section 3.1.4 "Multiple Sleep Commands" and Section 22.4.1 "Special Considerations for Using Two-Speed Start-up" for more details.

The FSCM will detect failures of the primary or secondary clock sources only. If the internal oscillator block fails, no failure would be detected, nor would any action be possible.

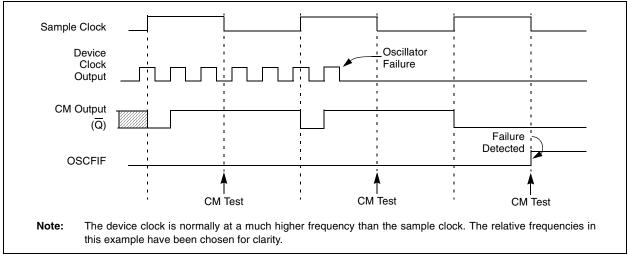
22.5.1 FSCM AND THE WATCHDOG TIMER

Both the FSCM and the WDT are clocked by the INTRC oscillator. Since the WDT operates with a separate divider and counter, disabling the WDT has no effect on the operation of the INTRC oscillator when the FSCM is enabled.

As already noted, the clock source is switched to the INTRC clock when a clock failure is detected; this may mean a substantial change in the speed of code execution. If the WDT is enabled with a small prescale value, a decrease in clock speed allows a WDT time-out to occur and a subsequent device Reset. For this reason, Fail-Safe Clock Monitor events also reset the WDT and postscaler, allowing it to start timing from when execution speed was changed and decreasing the likelihood of an erroneous time-out.

If the interrupt is disabled, subsequent interrupts while in Idle mode will cause the CPU to begin executing instructions while being clocked by the INTRC source.





22.5.2 EXITING FAIL-SAFE OPERATION

The fail-safe condition is terminated by either a device Reset or by entering a power-managed mode. On Reset, the controller starts the primary clock source specified in Configuration Register 2H (with any required start-up delays that are required for the oscillator mode, such as OST or PLL timer). The INTRC oscillator provides the device clock until the primary clock source becomes ready (similar to a Two-Speed Start-up). The clock source is then switched to the primary clock (indicated by the OSTS bit in the OSCCON register becoming set). The Fail-Safe Clock Monitor then resumes monitoring the peripheral clock.

The primary clock source may never become ready during start-up. In this case, operation is clocked by the INTOSC multiplexor. The OSCCON register will remain in its Reset state until a power-managed mode is entered.

22.5.3 FSCM INTERRUPTS IN POWER-MANAGED MODES

By entering a power-managed mode, the clock multiplexor selects the clock source selected by the OSCCON register. Fail-Safe Clock Monitoring of the power-managed clock source resumes in the power-managed mode.

If an oscillator failure occurs during power-managed operation, the subsequent events depend on whether or not the oscillator failure interrupt is enabled. If enabled (OSCFIF = 1), code execution will be clocked by the INTRC multiplexor. An automatic transition back to the failed clock source will not occur.

22.5.4 POR OR WAKE-UP FROM SLEEP

The FSCM is designed to detect oscillator failure at any point after the device has exited Power-on Reset (POR) or low-power Sleep mode. When the primary device clock is either EC or INTRC mode, monitoring can begin immediately following these events.

For HS or HSPLL modes, the situation is somewhat different. Since the oscillator may require a start-up time considerably longer than the FSCM sample clock time, a false clock failure may be detected. To prevent this, the internal oscillator block is automatically configured as the device clock and functions until the primary clock is stable (the OST and PLL timers have timed out). This is identical to Two-Speed Start-up mode. Once the primary clock is stable, the INTRC returns to its role as the FSCM source.

Note:	The same logic that prevents false oscillator failure interrupts on POR, or wake from Sleep, will also prevent the detection of the oscillator's failure to start at all following these events. This can be avoided by monitoring the OSTS bit and using a timing routine to determine if the oscillator is taking too long to start. Even so, no oscillator failure interrupt will be
	so, no oscillator failure interrupt will be flagged.

As noted in Section 22.4.1 "Special Considerations for Using Two-Speed Start-up", it is also possible to select another clock configuration and enter an alternate power-managed mode while waiting for the primary clock to become stable. When the new power-managed mode is selected, the primary clock is disabled.

22.6 Program Verification and Code Protection

For all devices in the PIC18F85J90 family of devices, the on-chip program memory space is treated as a single block. Code protection for this block is controlled by one Configuration bit, CP0. This bit inhibits external reads and writes to the program memory space. It has no direct effect in normal execution mode.

22.6.1 CONFIGURATION REGISTER PROTECTION

The Configuration registers are protected against untoward changes or reads in two ways. The primary protection is the write-once feature of the Configuration bits which prevents reconfiguration once the bit has been programmed during a power cycle. To safeguard against unpredictable events, Configuration bit changes resulting from individual cell-level disruptions (such as ESD events) will cause a parity error and trigger a device Reset.

The data for the Configuration registers is derived from the Flash Configuration Words in program memory. When the CP0 bit set, the source data for device configuration is also protected as a consequence.

22.7 In-Circuit Serial Programming

PIC18F85J90 family microcontrollers can be serially programmed while in the end application circuit. This is simply done with two lines for clock and data and three other lines for power, ground and the programming voltage. This allows customers to manufacture boards with unprogrammed devices and then program the microcontroller just before shipping the product. This also allows the most recent firmware or a custom firmware to be programmed.

22.8 In-Circuit Debugger

When the DEBUG Configuration bit is programmed to a '0', the In-Circuit Debugger functionality is enabled. This function allows simple debugging functions when used with MPLAB[®] IDE. When the microcontroller has this feature enabled, some resources are not available for general use. Table 22-4 shows which resources are required by the background debugger.

TABLE 22-4:	DEBUGGER RESOURCES
--------------------	--------------------

I/O pins:	RB6, RB7					
Stack:	2 levels					
Program Memory:	512 bytes					
Data Memory:	10 bytes					

23.0 INSTRUCTION SET SUMMARY

The PIC18F85J90 family of devices incorporate the standard set of 75 PIC18 core instructions, as well as an extended set of 8 new instructions for the optimization of code that is recursive or that utilizes a software stack. The extended set is discussed later in this section.

23.1 Standard Instruction Set

The standard PIC18 instruction set adds many enhancements to the previous PIC[®] MCU instruction sets, while maintaining an easy migration from these PIC MCU instruction sets. Most instructions are a single program memory word (16 bits), but there are four instructions that require two program memory locations.

Each single-word instruction is a 16-bit word divided into an opcode, which specifies the instruction type and one or more operands, which further specify the operation of the instruction.

The instruction set is highly orthogonal and is grouped into four basic categories:

- Byte-oriented operations
- Bit-oriented operations
- Literal operations
- Control operations

The PIC18 instruction set summary in Table 23-2 lists **byte-oriented**, **bit-oriented**, **literal** and **control** operations. Table 23-1 shows the opcode field descriptions.

Most byte-oriented instructions have three operands:

- 1. The file register (specified by 'f')
- 2. The destination of the result (specified by 'd')
- 3. The accessed memory (specified by 'a')

The file register designator 'f' specifies which file register is to be used by the instruction. The destination designator 'd' specifies where the result of the operation is to be placed. If 'd' is zero, the result is placed in the WREG register. If 'd' is one, the result is placed in the file register specified in the instruction.

All **bit-oriented** instructions have three operands:

- 1. The file register (specified by 'f')
- 2. The bit in the file register (specified by 'b')
- 3. The accessed memory (specified by 'a')

The bit field designator 'b' selects the number of the bit affected by the operation, while the file register designator 'f' represents the number of the file in which the bit is located. The **literal** instructions may use some of the following operands:

- A literal value to be loaded into a file register (specified by 'k')
- The desired FSR register to load the literal value into (specified by 'f')
- No operand required (specified by '—')

The **control** instructions may use some of the following operands:

- A program memory address (specified by 'n')
- The mode of the CALL or RETURN instructions (specified by 's')
- The mode of the table read and table write instructions (specified by 'm')
- No operand required (specified by '—')

All instructions are a single word, except for four double-word instructions. These instructions were made double-word to contain the required information in 32 bits. In the second word, the 4 MSbs are '1's. If this second word is executed as an instruction (by itself), it will execute as a NOP.

All single-word instructions are executed in a single instruction cycle, unless a conditional test is true or the program counter is changed as a result of the instruction. In these cases, the execution takes two instruction cycles with the additional instruction cycle(s) executed as a NOP.

The double-word instructions execute in two instruction cycles.

One instruction cycle consists of four oscillator periods. Thus, for an oscillator frequency of 4 MHz, the normal instruction execution time is 1 μ s. If a conditional test is true, or the program counter is changed as a result of an instruction, the instruction execution time is 2 μ s. Two-word branch instructions (if true) would take 3 μ s.

Figure 23-1 shows the general formats that the instructions can have. All examples use the convention 'nnh' to represent a hexadecimal number.

The Instruction Set Summary, shown in Table 23-2, lists the standard instructions recognized by the Microchip MPASM[™] Assembler.

Section 23.1.1 "Standard Instruction Set" provides a description of each instruction.

TABLE 23-1: OPCODE FIELD DESCRIPTIONS

Field	Description
a	BAM access bit:
~	a = 0: RAM location in Access RAM (BSR register is ignored)
	a = 1: RAM bank is specified by BSR register
bbb	Bit address within an 8-bit file register (0 to 7).
BSR	Bank Select Register. Used to select the current RAM bank.
C, DC, Z, OV, N	ALU Status bits: Carry, Digit Carry, Zero, Overflow, Negative.
d	Destination select bit:
	d = 0: store result in WREG
	d = 1: store result in file register f
dest	Destination: either the WREG register or the specified register file location.
f	8-bit Register file address (00h to FFh), or 2-bit FSR designator (0h to 3h).
fs	12-bit Register file address (000h to FFFh). This is the source address.
f _d	12-bit Register file address (000h to FFFh). This is the destination address.
GIE	Global Interrupt Enable bit.
k	Literal field, constant data or label (may be either an 8-bit, 12-bit or a 20-bit value).
label	Label name.
mm	The mode of the TBLPTR register for the table read and table write instructions. Only used with table read and table write instructions:
*	No Change to register (such as TBLPTR with table reads and writes)
*+	Post-Increment register (such as TBLPTR with table reads and writes)
*_	Post-Decrement register (such as TBLPTR with table reads and writes)
+*	Pre-Increment register (such as TBLPTR with table reads and writes)
n	The relative address (2's complement number) for relative branch instructions or the direct address for
	Call/Branch and Return instructions.
PC	Program Counter.
PCL	Program Counter Low Byte.
PCH	Program Counter High Byte.
PCLATH	Program Counter High Byte Latch.
PCLATU	Program Counter Upper Byte Latch.
PD	Power-Down bit.
PRODH	Product of Multiply High Byte.
PRODL	Product of Multiply Low Byte.
s	Fast Call/Return mode select bit:
	s = 0: do not update into/from shadow registers
	s = 1: certain registers loaded into/from shadow registers (Fast mode)
TBLPTR	21-bit Table Pointer (points to a Program Memory location).
TABLAT	8-bit Table Latch.
TO	Time-out bit.
TOS	Top-of-Stack.
u	Unused or Unchanged.
WDT	Watchdog Timer.
WREG	Working register (accumulator).
x	Don't care ('0' or '1'). The assembler will generate code with $x = 0$. It is the recommended form of use for compatibility with all Microchip software tools.
Zs	7-bit offset value for Indirect Addressing of register files (source).
z _d	7-bit offset value for Indirect Addressing of register files (destination).
{ }	Optional argument.
[text]	Indicates an Indexed Address.
(text)	The contents of text.
[expr] <n></n>	Specifies bit n of the register indicated by the pointer $expr$.
\rightarrow	Assigned to.
< >	Register bit field.
E	In the set of.

Byte-oriented file register operations	Example Instruction
15 10 9 8 7 0	
OPCODE d a f (FILE #)	ADDWF MYREG, W, B
$ d = 0 \ \text{for result destination to be WREG register} \\ d = 1 \ \text{for result destination to be file register (f)} \\ a = 0 \ \text{to force Access Bank} \\ a = 1 \ \text{for BSR to select bank} \\ f = 8 \ \text{bit file register address} $	
Byte to Byte move operations (2-word)	
15 12 11 0	
OPCODE f (Source FILE #)	MOVFF MYREG1, MYREG2
<u>15 12 11 0</u>	
1111 f (Destination FILE #)	
f = 12-bit file register address	
Bit-oriented file register operations	
15 12 11 9 8 7 0	
OPCODE b (BIT #) a f (FILE #)	BSF MYREG, bit, B
 b = 3-bit position of bit in file register (f) a = 0 to force Access Bank a = 1 for BSR to select bank f = 8-bit file register address 	
Literal operations	
15 8 7 0	
OPCODE k (literal)	MOVLW 7Fh
k = 8-bit immediate value	
Control operations	
CALL, GOTO and Branch operations	
15 8 7 0 OPCODE n<7:0> (literal)	GOTO Label
	GOIO HADEI
15 12 11 0 1111 n<19:8> (literal)	
n = 20-bit immediate value	
15 8 7 0	
OPCODE S n<7:0> (literal)	CALL MYFUNC
15 12 11 0	
1111 n<19:8> (literal)	
S = Fast bit	
15 11 10 0	
OPCODE n<10:0> (literal)	BRA MYFUNC
15 8 7 0 OPCODE n<7:0> (literal)	BC MYFUNC

TABLE 23-2:PIC18F85J90 FAMILY INSTRUCTION SET

OperandsBYTE-ORIENTED OPERATIONADDWFf, d, aAdd WRECADDWFCf, d, aAdd WRECANDWFf, d, aAdd WRECCLRFf, aClear fCOMFf, d, aComplemeCPFSEQf, aCompare fCPFSGTf, aCompare fDECFf, d, aDecrementDECFSZf, d, aDecrementINCFf, d, aIncrementINCFSZf, d, aIncrementINFSNZf, d, aIncrementIORWFf, d, aIncrementMOVFf, d, aMove fMOVFFfs, fdMove fs (sMOVFFfs, fdMove fs (s	G and f G and Carry bit to f	Cycles	MSb			LSb	Affected	Notes
ADDWFf, d, aAdd WRECADDWFCf, d, aAdd WRECANDWFf, d, aAND WRECCLRFf, aClear fCOMFf, d, aComplemeCPFSEQf, aCompare fCPFSGTf, aCompare fDECFf, d, aDecrementDECFSZf, d, aDecrementINCFf, d, aIncrementINCFSZf, d, aIncrementINFSNZf, d, aIncrementINFSNZf, d, aIncrementINCFSZf, d, aIncrementINCFSZf, d, aIncrementINCFSZf, d, aIncrementINCFSZf, d, aIncrementINCFSZf, d, aIncrementINFSNZf, d, aIncrementINCFSTf, d, aIncrementINCFSTf, d, aInclusive OMOVFf, d, aMove fMOVFFfs, fdMove fs (sfd (dfd (d)	G and f G and Carry bit to f	1	0010					
ADDWFCf, d, aAdd WRECANDWFf, d, aAND WRECCLRFf, aClear fCOMFf, d, aComplemeCPFSEQf, aCompare fCPFSGTf, aCompare fDECFf, d, aDecrementDECFSZf, d, aDecrementDCFSNZf, d, aDecrementINCFf, d, aIncrementINCFSZf, d, aIncrementINCFSZf, d, aIncrementINCFSZf, d, aIncrementINFSNZf, d, aIncrementINFSNZf, d, aIncrementINCFSZf, d, aIncrementINFSNZf, d, aIncrementINFSNZf, d, aIncrementIORWFf, d, aMove fMOVFFf, d, aMove fMOVFFf, d, aMove ff, d, afd	G and Carry bit to f	1	0010	_				
ANDWFf, d, aAND WRECLRFf, aClear fCOMFf, d, aComplemeCPFSEQf, aCompare fCPFSGTf, aCompare fDECFf, d, aDecrementDECFSZf, d, aDecrementDCFSNZf, d, aDecrementINCFf, d, aIncrementINCFSZf, d, aIncrementINCFSZf, d, aIncrementINFSNZf, d, aInclusive OMOVFf, d, aMove fMOVFFfs, fdMove fs (sfd (dfd (d)		1		01da	ffff	ffff	C, DC, Z, OV, N	1, 2
$\begin{array}{cccc} \text{CLRF} & \text{f, a} & \text{Clear f} \\ \text{COMF} & \text{f, d, a} & \text{Compleme} \\ \text{CPFSEQ} & \text{f, a} & \text{Compare f} \\ \text{CPFSGT} & \text{f, a} & \text{Compare f} \\ \text{CPFSLT} & \text{f, a} & \text{Compare f} \\ \text{DECF} & \text{f, d, a} & \text{Decrement} \\ \text{DECFSZ} & \text{f, d, a} & \text{Decrement} \\ \text{DCFSNZ} & \text{f, d, a} & \text{Decrement} \\ \text{INCF} & \text{f, d, a} & \text{Increment} \\ \text{INCFSZ} & \text{f, d, a} & \text{Increment} \\ \text{INCFSZ} & \text{f, d, a} & \text{Increment} \\ \text{INCFSZ} & \text{f, d, a} & \text{Increment} \\ \text{INFSNZ} & \text{Increment} \\ \text{Increment} \\$	G with f	•	0010	00da	ffff	ffff	C, DC, Z, OV, N	1, 2
COMF f, d, a Compleme CPFSEQ f, a Compare f CPFSGT f, a Compare f CPFSLT f, a Compare f DECF f, d, a Decrement DECFSZ f, d, a Decrement INCF f, d, a Increment INCFSZ f, d, a Increment INFSNZ f, d, a Increment INFSNZ f, d, a Increment INFSNZ f, d, a Increment MOVF f, d, a Move f MOVF f, d, a Move f MOVFF f _s , f _d Move f _s (s		1	0001	01da	ffff	ffff	Z, N	1,2
CPFSEQ f, a Compare f CPFSGT f, a Compare f CPFSLT f, a Compare f DECF f, d, a Decrement DECFSZ f, d, a Decrement INCF f, d, a Increment INCFSZ f, d, a Increment INFSNZ f, d, a Increment INFSNZ f, d, a Increment INFSNZ f, d, a Increment MOVF f, d, a Move f MOVF f, d, a Move f MOVFF f _s , f _d Move f _s (s		1	0110	101a	ffff	ffff	Z	2
CPFSGT f, a Compare f CPFSLT f, a Compare f DECF f, d, a Decrement DECFSZ f, d, a Decrement INCF f, d, a Decrement INCFSZ f, d, a Increment INFSNZ f, d, a Increment INFSNZ f, d, a Increment INFSNZ f, d, a Increment INFSNZ f, d, a Increment MOVF f, d, a Move f MOVF f, d, a Move f MOVFF f _s , f _d Move f _s (s	ent f	1	0001	11da	ffff	ffff	Z, N	1, 2
CPFSLT f, a Compare f DECF f, d, a Decrement DECFSZ f, d, a Decrement DCFSNZ f, d, a Decrement INCF f, d, a Increment INFSNZ f, d, a Increment INFSNZ f, d, a Increment INFSNZ f, d, a Increment INFSNZ f, d, a Increment MOVF f, d, a Move f MOVF f, d, a Move f MOVFF f _s , f _d Move f _s (s	with WREG, skip =	1 (2 or 3)	0110	001a	ffff	ffff	None	4
$\begin{array}{c cccc} DECF & f, d, a & Decrement \\ DECFSZ & f, d, a & Decrement \\ DCFSNZ & f, d, a & Decrement \\ INCF & f, d, a & Increment \\ INCFSZ & f, d, a & Increment \\ INFSNZ & f, d, a & Increment \\ IORWF & f, d, a & Inclusive \\ MOVFF & f, d, a & Move f \\ MOVFF & f_{s}, f_{d} & Move f_{s} (s \\ f_{d} (d \\ d \\ f_{d} (d \\ d \\ d \\ f_{d} (d \\ d \\ d \\ f_{d} (d \\ d \\ $	with WREG, skip >	1 (2 or 3)	0110	010a	ffff	ffff	None	4
DECFSZ f, d, a Decrement DCFSNZ f, d, a Decrement INCF f, d, a Increment INCFSZ f, d, a Increment INFSNZ f, d, a Increment IORWF f, d, a Inclusive O MOVF f, d, a Move f MOVFF f _s , f _d Move f _s (s f _d (d	with WREG, skip <	1 (2 or 3)	0110	000a	ffff	ffff	None	1, 2
DCFSNZ f, d, a Decrement INCF f, d, a Increment INCFSZ f, d, a Increment INFSNZ f, d, a Increment IORWF f, d, a Inclusive O MOVF f, d, a Move f MOVFF f _s , f _d Move f _s (s f _d (d	tf	1 1	0000	01da	ffff	ffff	C, DC, Z, OV, N	1, 2, 3, 4
DCFSNZf, d, aDecrementINCFf, d, aIncrementINCFSZf, d, aIncrementINFSNZf, d, aIncrementIORWFf, d, aInclusive OMOVFf, d, aMove fMOVFFf_s, f_dMove f_s (s)f_d (d)f_d (d)	t f, Skip if 0	1 (2 or 3)	0010	11da	ffff	ffff	None	1, 2, 3, 4
$\begin{array}{llllllllllllllllllllllllllllllllllll$	t f, Skip if Not 0	1 (2 or 3)	0100	11da	ffff	ffff	None	1, 2
$\begin{array}{llllllllllllllllllllllllllllllllllll$		1 ΄	0010	10da	ffff	ffff	C, DC, Z, OV, N	1, 2, 3, 4
$\begin{array}{c c} INFSNZ & f, d, a & Increment\\ IORWF & f, d, a & Inclusive \\ MOVF & f, d, a & Move \\ MOVFF & f_{s}, f_{d} & Move \\ f_{d} (d \\ f_{d} (d \\ d \\ f_{d} (d \\ d \\ \mathsf$	f, Skip if 0	1 (2 or 3)	0011	11da	ffff		None	4
$\begin{array}{llllllllllllllllllllllllllllllllllll$	f, Skip if Not 0	1 (2 or 3)	0100	10da	ffff	ffff	None	1, 2
$\begin{array}{llllllllllllllllllllllllllllllllllll$	R WREG with f	1 ΄	0001	00da	ffff	ffff	Z, N	1, 2
$\begin{array}{llllllllllllllllllllllllllllllllllll$		1	0101	00da	ffff	ffff	Z, N	1
f _d (d	ource) to 1st word	2	1100	ffff	ffff	ffff	None	
	lestination) 2nd word		1111	ffff	ffff	ffff		
MOVWF f, a Move WRE		1	0110	111a	ffff	ffff	None	
MULWF f, a Multiply Wi	REG with f	1	0000	001a	ffff	ffff	None	1, 2
NEGF f, a Negate f		1	0110	110a	ffff	ffff	C, DC, Z, OV, N	<i>`</i>
	t f through Carry	1	0011	01da	ffff		C, Z, N	1, 2
	t f (No Carry)	1	0100	01da	ffff	ffff		,
	ht f through Carry	1	0011	00da	ffff		C, Z, N	
, , j	ht f (No Carry)	1	0100	00da	ffff	ffff		
SETF f, a Set f	(),	1	0110	100a	ffff	ffff	None	1, 2
·	from WREG with	1	0101	01da	ffff	ffff	C, DC, Z, OV, N	,
borrow								
	/REG from f	1	0101	11da	ffff	ffff	C, DC, Z, OV, N	1, 2
	/REG from f with	1		10da	ffff		C, DC, Z, OV, N	,
borrow						_	, _, ,, _	
SWAPF f, d, a Swap nibbl	les in f	1	0011	10da	ffff	ffff	None	4
TSTFSZ f, a Test f, skip								1, 2
XORWF f, d, a Exclusive C		1 (2 or 3)	0110	011a	ffff	ffff	None	1.2

Note 1: When a PORT register is modified as a function of itself (e.g., MOVF PORTB, 1, 0), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and, where applicable, d = 1), the prescaler will be cleared if assigned.

3: If Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

4: Some instructions are two-word instructions. The second word of these instructions will be executed as a NOP unless the first word of the instruction retrieves the information embedded in these 16-bits. This ensures that all program memory locations have a valid instruction.

Mnemonic, Operands				16-bit Instruction Word				Status	
		Description	Cycles	MSb			LSb	Affected	Notes
BIT-ORIE	NTED OP	ERATIONS							
BCF	f, b, a	Bit Clear f	1	1001	bbba	ffff	ffff	None	1, 2
BSF	f, b, a	Bit Set f	1	1000	bbba	ffff	ffff	None	1, 2
BTFSC	f, b, a	Bit Test f, Skip if Clear	1 (2 or 3)	1011	bbba	ffff	ffff	None	3, 4
BTFSS	f, b, a	Bit Test f, Skip if Set	1 (2 or 3)	1010	bbba	ffff	ffff	None	3, 4
BTG	f, b, a	Bit Toggle f	1	0111	bbba	ffff	ffff	None	1, 2
CONTRO	L OPERA	TIONS							
BC	n	Branch if Carry	1 (2)	1110	0010	nnnn	nnnn	None	
BN	n	Branch if Negative	1 (2)	1110	0110	nnnn	nnnn	None	
BNC	n	Branch if Not Carry	1 (2)	1110	0011	nnnn	nnnn	None	
BNN	n	Branch if Not Negative	1 (2)	1110	0111	nnnn	nnnn		
BNOV	n	Branch if Not Overflow	1 (2)	1110	0101	nnnn		None	
BNZ	n	Branch if Not Zero	1 (2)	1110	0001	nnnn		None	
BOV	n	Branch if Overflow	1 (2)	1110	0100	nnnn	nnnn		
BRA	n	Branch Unconditionally	2	1101	0nnn	nnnn	nnnn		
BZ	n	Branch if Zero	1 (2)	1110	0000	nnnn	nnnn	None	
CALL	n, s	Call subroutine 1st word	2	1110	110s	kkkk	kkkk	None	
		2nd word			kkkk	kkkk	kkkk		
CLRWDT	—	Clear Watchdog Timer	1	0000	0000	0000	0100	TO, PD	
DAW	—	Decimal Adjust WREG	1	0000	0000	0000	0111	С	
GOTO	n	Go to address 1st word	2	1110	1111	kkkk	kkkk	None	
		2nd word			kkkk	kkkk	kkkk		
NOP	—	No Operation	1	0000	0000	0000	0000		
NOP	—	No Operation	1		XXXX	XXXX		None	4
POP		Pop top of return stack (TOS)	1	0000	0000	0000		None	
PUSH	—	Push top of return stack (TOS)	1	0000	0000	0000		None	
RCALL	n	Relative Call	2	1101	1nnn	nnnn	nnnn		
RESET		Software device Reset	1	0000	0000	1111	1111		
RETFIE	S	Return from interrupt enable	2	0000	0000	0001		GIE/GIEH, PEIE/GIEL	
RETLW	k	Return with literal in WREG	2	0000	1100	kkkk	kkkk	None	
RETURN	S	Return from Subroutine	2 1	0000	0000	0001	001s	None	
SLEEP		Go into Standby mode	1	0000	0000	0000	0011	TO, PD	

TABLE 23-2:	PIC18F85J90 FAMILY INSTRUCTION SET ((CONTINUED))
			,

Note 1: When a PORT register is modified as a function of itself (e.g., MOVF PORTB, 1, 0), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and, where applicable, d = 1), the prescaler will be cleared if assigned.

3: If Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

4: Some instructions are two-word instructions. The second word of these instructions will be executed as a NOP unless the first word of the instruction retrieves the information embedded in these 16-bits. This ensures that all program memory locations have a valid instruction.

Mnem				16-bit Instruction Word				Status	
Mnemonic, Operands		Description	Cycles	MSb			LSb	Affected	Notes
LITERAL C		•							
ADDLW	k	Add literal and WREG	1	0000	1111	kkkk	kkkk	C, DC, Z, OV, N	
ANDLW	k	AND literal with WREG	1	0000	1011	kkkk	kkkk	Z, N	
IORLW	k	Inclusive OR literal with WREG	1	0000	1001	kkkk	kkkk	Z, N	
LFSR	f, k	Move literal (12-bit) 2nd word	2	1110	1110	00ff	kkkk	None	
		to FSR(f) 1st word		1111	0000	kkkk	kkkk		
MOVLB	k	Move literal to BSR<3:0>	1	0000	0001	0000	kkkk	None	
MOVLW	k	Move literal to WREG	1	0000	1110	kkkk	kkkk	None	
MULLW	k	Multiply literal with WREG	1	0000	1101	kkkk	kkkk	None	
RETLW	k	Return with literal in WREG	2	0000	1100	kkkk	kkkk	None	
SUBLW	k	Subtract WREG from literal	1	0000	1000	kkkk	kkkk	C, DC, Z, OV, N	
XORLW	k	Exclusive OR literal with WREG	1	0000	1010	kkkk	kkkk	Z, N	
DATA MEN	MORY ←	> PROGRAM MEMORY OPERATI	ONS						
TBLRD*		Table Read	2	0000	0000	0000	1000	None	
TBLRD*+		Table Read with post-increment		0000	0000	0000	1001	None	
TBLRD*-		Table Read with post-decrement		0000	0000	0000	1010	None	
TBLRD+*		Table Read with pre-increment		0000	0000	0000	1011	None	
TBLWT*		Table Write	2	0000	0000	0000	1100	None	
TBLWT*+		Table Write with post-increment		0000	0000	0000	1101	None	
TBLWT*-		Table Write with post-decrement		0000	0000	0000	1110	None	
TBLWT+*		Table Write with pre-increment		0000	0000	0000	1111	None	

TABLE 23-2: PIC18F85J90 FAMILY INSTRUCTION SET (CONTINUED)

Note 1: When a PORT register is modified as a function of itself (e.g., MOVF PORTB, 1, 0), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and, where applicable, d = 1), the prescaler will be cleared if assigned.

3: If Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

4: Some instructions are two-word instructions. The second word of these instructions will be executed as a NOP unless the first word of the instruction retrieves the information embedded in these 16-bits. This ensures that all program memory locations have a valid instruction.

f {,d {,a}}

01da

ffff

ffff

23.1.1 STANDARD INSTRUCTION SET

ADDLW ADD Literal to W			ADDWF	ADD W to	f	
Syntax:	ADDLW	k		Syntax:	ADDWF	f {,d {
Operands:	$0 \le k \le 255$			Operands:	$0 \le f \le 255$	
Operation:	$(W) + k \rightarrow V$	W			d ∈ [0,1] a ∈ [0,1]	
Status Affected:	N, OV, C, D	0C, Z		Operation:	a ∈ [0,1] (W) + (f) →	doct
Encoding:	0000	1111 kkł	ck kkkk	Status Affected:	(₩) + (I) → N, OV, C, E	
Description:		ts of W are ad k' and the res		Encoding: Description:	0010 Add W to re	01da
Words:	1			•	result is sto	
Cycles:	1				result is sto (default).	ored ba
Q Cycle Activity:					If 'a' is '0', t	the Ac
Q1	Q2	Q3	Q4		lf 'a' is '1', t	he BS
Decode	Read literal 'k'	Process Data	Write to W		GPR bank If 'a' is '0' a	•
<u>Example:</u> Before Instruc W = After Instructi	ction 10h	.5h			set is enab in Indexed mode wher Section 23 Bit-Oriente Literal Offs	Literal never f 3.2.3 "I ed Ins
W =	25h			Words:	1	
				Cycles:	1	
				Q Cycle Activity:		
				Q1	Q2	
				Decode	Read register 'f'	Pro D
				Example:	ADDWF	REG,
				Before Instruc W REG	ction = 17h = 0C2h	
				After Instructi		
				W_	= 0D9h	

dd W to register 'f'. If 'd' is '0', the sult is stored in W. If 'd' is '1', the sult is stored back in register 'f' efault). 'a' is '0', the Access Bank is selected. 'a' is '1', the BSR is used to select the PR bank (default). 'a' is '0' and the extended instruction et is enabled, this instruction operates Indexed Literal Offset Addressing ode whenever $f \leq 95$ (5Fh). See ection 23.2.3 "Byte-Oriented and t-Oriented Instructions in Indexed teral Offset Mode" for details. Q2 Q3 Q4 Read Process Write to ister 'f' Data destination DDWF REG, 0, 0 17h 0C2h 0D9h REG = 0C2h

Note: All PIC18 instructions may take an optional label argument preceding the instruction mnemonic for use in symbolic addressing. If a label is used, the instruction format then becomes: {label} instruction argument(s).

ADDWFC	ADD W and	d Carry b	it to f				
Syntax:	ADDWFC	f {,d {,a	}}				
Operands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$						
Operation:	(W) + (f) +	$(C) \rightarrow des$	st				
Status Affected:	N,OV, C, D	C, Z					
Encoding:	0010	0010 00da ffff ffff					
Description:	Add W, the location 'f'. placed in W placed in d	If 'd' is 'o' /. If 'd' is '	, the res 1', the re	ult is sult is			
	If 'a' is '0', t If 'a' is '1', t GPR bank	he BSR is					
	If 'a' is '0' and the extended instructio set is enabled, this instruction operate in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 23.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexe Literal Offset Mode" for details.						
Words:	1						
Cycles:	1						
Q Cycle Activity:							
Q1	Q2	Q3		Q4			
Decode	Read register 'f'	Proces Data		Vrite to stination			
Example:	ADDWFC	REG,	0, 1				
Before Instruc Carry bit REG W	= 1 = 02h = 4Dh						
After Instructic Carry bit REG W							

ANDLW	AND Litera	al with W	'					
Syntax:	ANDLW	k						
Operands:	$0 \le k \le 255$							
Operation:	(W) .AND.	(W) .AND. $k \rightarrow W$						
Status Affected:	N, Z	N, Z						
Encoding:	0000	1011	kkk	k	kkkk			
Description:	The conten 8-bit literal							
Words:	1							
Cycles:	1							
Q Cycle Activity:								
Q1	Q2	Q3			Q4			
Decode	Read literal 'k'	Proce Data		N	/rite to W			
Example:	ANDLW	05Fh						
Before Instruc W After Instructio	= A3h on							
W	= 03h							

ANDWF	AND W wit	th f		BC		Branch if C	Carry		
Syntax:	ANDWF	f {,d {,a}}		Synt	ax:	BC n			
Operands:	$0 \le f \le 255$			Oper	ands:	-128 ≤ n ≤ ⁻	127		
	d ∈ [0,1] a ∈ [0,1]			Oper	ation:	,	if Carry bit is '1', (PC) + 2 + 2n \rightarrow PC		
Operation:	(W) .AND.	(f) \rightarrow dest		Statu	s Affected:	None			
Status Affected:	N, Z			Enco	oding:	1110	0010 nn:	nn nnnn	
Encoding:	0001	0001 01da ffff ffff			ription:	If the Carry	bit is '1'. then	the program	
Description: The contents of W are ANDed with register 'f'. If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f' (default).					If the Carry bit is '1', then the program will branch. The 2's complement number '2n' is added to the PC. Since the PC will have				
	If 'a' is 'o', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).					incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a two-cycle instruction.			
	lf 'a' is '0' a	and the extend	ed instruction	Word	ls:	1			
			ction operates	Cycl	es:	1(2)			
	mode wher	Literal Offset <i>i</i> never f ≤ 95 (5 3.2.3 " Byte-O r	Fh). See		ycle Activity: imp:				
		ed Instruction			Q1	Q2	Q3	Q4	
	Literal Off	set Mode" for	details.		Decode	Read literal	Process	Write to	
Words:	1				Nia	ʻn' No	Data	PC	
Cycles:	1				No operation	operation	No operation	No operation	
Q Cycle Activity:				If N	o Jump:				
Q1	Q2	Q3	Q4		Q1	Q2	Q3	Q4	
Decode	Read register 'f'	Process Data	Write to destination		Decode	Read literal	Process	No	
	register i	Data	destination			'n'	Data	operation	
Example:	ANDWF	REG, 0, 0	1	Exar	nple:	HERE	BC 5		
Before Instruc W REG After Instructi	= 17h = C2h				Before Instruction PC After Instruction	= ad	dress (HERE)	
W REG	= 02h = C2h				If Carry PC If Carry PC	= 0;	dress (HERE dress (HERE		

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BCF	Bit Clear f			BN		Branch if N	legative			
Syntax:	BCF f, b	{,a}		Synt	ax:	BN n				
Operands:	$0 \leq f \leq 255$			Oper	ands:	-128 ≤ n ≤ ⁻	127			
	0 ≤ b ≤ 7 a ∈ [0,1]			Oper	ation:	0	if Negative bit is '1', (PC) + 2 + 2n \rightarrow PC			
Operation:	$0 \rightarrow f < b >$			Statu	is Affected:	None				
Status Affected:	None		Enco	oding:	1110	0110 nn:	nn nnnn			
Encoding:	1001	1001 bbba ffff ffff			ription:		tive bit is '1', tl			
Description:	Bit 'b' in reg	gister 'f' is cle	eared.	2000	inpuoli.	program wi				
	If 'a' is 'o', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).				The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next					
If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing				instruction, the new address will be PC + 2 + 2n. This instruction is then a two-cycle instruction.						
	mode when	ever f ≤ 95	(5Fh). See	Word	ls:	1				
		•	Driented and ons in Indexed	Cycle	es:	1(2)				
		set Mode" fo		Q C If Ju	ycle Activity:					
Words:	1			1100	Q1	Q2	Q3	Q4		
Cycles:	1				Decode	Read literal	Process	Write to		
Q Cycle Activity:						'n'	Data	PC		
Q1	Q2	Q3	Q4	I	No	No	No	No		
Decode	Read	Process	Write	IF NI.	operation	operation	operation	operation		
	register 'f'	Data	register 'f'		o Jump: Q1	Q2	Q3	Q4		
Example:	BCF F	LAG REG,	7, 0		Decode	Read literal	Process	No		
Before Instruct		,	., .			'n'	Data	operation		
	EG = C7h									
After Instruction				<u>Exar</u>	nple:	HERE	BN Jump			
FLAG_R	EG = 47h				Before Instruc PC After Instructi If Negati	= ad on ve = 1;	dress (HERE)			
					PC If Negati PC	ve = 0;	dress (Jump) dress (HERE			

BNC	Branch if N	lot Carry		BNN		Branch if I	Not Negative	
Syntax:	BNC n			Synta	ax:	BNN n	BNN n	
Operands:	-128 ≤ n ≤ 1	127		Oper	ands:	-128 ≤ n ≤	127	
Operation:	if Carry bit i (PC) + 2 + 2			Oper	ation:	if Negative bit is '0', (PC) + 2 + 2n \rightarrow PC		
Status Affected:	None			Statu	s Affected:	None		
Encoding:	1110	0011 nn	nn nnnn	Enco	ding:	1110	0111 nn	inn nnnn
Description:	If the Carry bit is '0', then the program will branch.				ription:	If the Nega program wi	tive bit is '0', t Il branch.	hen the
	added to the incremented instruction,	d to fetch the r the new addre n. This instruct	e PC will have next ess will be			added to th incremente instruction,	d to fetch the the new addr n. This instruc	ne PC will have next ress will be
Words:	1			Word	s:	1		
Cycles:	1(2)			Cycle	es:	1(2)		
Q Cycle Activity: If Jump:				Q C If Ju	ycle Activity: mp:			
Q1	Q2	Q3	Q4		Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	Write to PC		Decode	Read literal 'n'	Process Data	Write to PC
No	No	No	No		No	No	No	No
operation	operation	operation	operation		operation	operation	operation	operation
If No Jump:				lf No	Jump:			
Q1	Q2	Q3	Q4		Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	No operation		Decode	Read literal 'n'	Process Data	No operation
Example:	HERE	BNC Jump		Exan	nple:	HERE	BNN Jum <u>r</u>)
Before Instruc PC After Instructi If Carry PC	= adv on = 0;	dress (HERE)		Before Instruc PC After Instructi If Negati PC	= ad on ive = 0;	dress (HERE	
If Carry PC	= 1;	dress (Jump) dress (HERE	+ 2)		If Negati PC	ve = 1;	dress (Jump dress (HERE	

BNO	v	Branch if N	Not Overflow		BNZ		Branch if I	lot Zero		
Synta	ax:	BNOV n			Synta	ax:	BNZ n			
Oper	ands:	-128 ≤ n ≤ ⁻	127		Oper	ands:	-128 ≤ n ≤ 1	127		
Oper	ration:	if Overflow (PC) + 2 + 2	,		Oper	ation:		if Zero bit is '0', (PC) + 2 + 2n \rightarrow PC		
Statu	is Affected:	None			Statu	s Affected:	None			
Enco	oding:	1110 0101 nnnn nnnn		Enco	ding:	1110	1110 0001 nnnn			
Desc	cription:	If the Overf program wi	low bit is '0', th Il branch.	nen the	Desc	ription:	If the Zero will branch.	bit is '0', then	the program	
		added to the incremente instruction,	nplement num e PC. Since the d to fetch the r the new addre n. This instruct nstruction.	e PC will have next ess will be			added to th incremente instruction,	d to fetch the the new add n. This instruc	he PC will have next	
Word	ds:	1			Word	ls:	1			
Cycle	es:	1(2)			Cycle	es:	1(2)			
	ycle Activity: imp:				Q C If Ju	ycle Activity: mp:				
	Q1	Q2	Q3	Q4		Q1	Q2	Q3	Q4	
	Decode	Read literal 'n'	Process Data	Write to PC		Decode	Read literal 'n'	Process Data	Write to PC	
	No	No	No	No		No	No	No	No	
IF NL	operation	operation	operation	operation	If NL	operation	operation	operation	operation	
	Q1	Q2	Q3	Q4		Q1	Q2	Q3	Q4	
	Decode	Read literal	Process	No		Decode	Read literal	Process	No	
	200040	ʻn'	Data	operation		200040	'n'	Data	operation	
<u>Exan</u>	Before Instruc PC After Instructio	= ad	BNOV Jump dress (HERE)	1	<u>Exan</u>	Before Instruc PC After Instructi	= ad	BNZ Jumj dress (HERE	-	
	If Overflo PC If Overflo PC	= ad w = 1;	dress (Jump) dress (HERE			If Zero PC If Zero PC	= 1;	dress (Jump dress (HERE		

BRA		Unconditio	Unconditional Branch							
Synta	ax:	BRA n								
Oper	ands:	-1024 ≤ n ≤	≤ 1023							
Oper	ation:	(PC) + 2 +	$(PC) + 2 + 2n \rightarrow PC$							
Statu	s Affected:	None	None							
Enco	ding:	1101	0nnn	nnnn	nnnn					
Desc	ription:	Add the 2's the PC. Sir incremente instruction, PC + 2 + 2 two-cycle in	nce the P ed to fetch the new n. This in	C will have the next address struction	re will be					
Word	ls:	1								
Cycle	es:	2								
QC	ycle Activity:									
	Q1	Q2	Q3		Q4					
	Decode	Read literal 'n'	Proce Data		Vrite to PC					
	No operation	No operation	No operat	ion o	No peration					
	n <u>ple:</u> Before Instruc PC After Instructio PC	= ac	Idress (1	Jump HERE) Jump)						

BSF	Bit Set f							
Syntax:	BSF f, b	{,a}						
Operands:	0 ≤ f ≤ 255 0 ≤ b ≤ 7 a ∈ [0,1]							
Operation:	$\texttt{1} \rightarrow f{}}$							
Status Affected:	None	None						
Encoding:	1000	bbba	ffff	ffff				
Description:	Bit 'b' in reg	Bit 'b' in register 'f' is set.						
	lf 'a' is '1', t	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).						
	set is enab in Indexed mode when Section 23 Bit-Oriente	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 23.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.						
Words:	1							
Cycles:	1							
Q Cycle Activity:								
Q1	Q2	Q3		Q4				
Decode	Read register 'f'	Proce Data		Write gister 'f'				

Decode	Read	Process	Wr
	register 'f'	Data	regist

FLAG_REG, 7, 1

Example:

Before Instruction FLAG_REG = 0Ah After Instruction FLAG_REG = 8Ah

BSF

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BTFSC	Bit Test File	, Skip if Clear		BTFS	S	Bit Test File	, Skip if Set		
Syntax:	BTFSC f, b	{,a}		Synta:	x:	BTFSS f, b {	,a}		
Operands:	0 ≤ f ≤ 255 0 ≤ b ≤ 7 a ∈ [0,1]			Opera	inds:	0 ≤ f ≤ 255 0 ≤ b < 7 a ∈ [0,1]			
Operation:	skip if (f)	= 0		Opera	ition:	skip if (f)	= 1		
Status Affected:	None			Status	Affected:	None			
Encoding:	1011	bbba ff	ff ffff	Encod	ling:	1010	1010 bbba ffff :		
Description: If bit 'b' in register 'f' is '0', then the next instruction is skipped. If bit 'b' is '0', then the next instruction fetched during the current instruction execution is discarded and a NOP is executed instead, making this a two-cycle instruction.		Descr	iption:	If bit 'b' in register 'f' is '1', then the next instruction is skipped. If bit 'b' is '1', then the next instruction fetched during the current instruction execution is discarded and a NOP is executed instead, making this a two-cycle instruction.					
		BSR is used to	k is selected. If a select the				e Access Bank BSR is used to lefault).		
	is enabled, th Indexed Lite whenever f ≤ Section 23.2 Bit-Oriented	d the extended his instruction ral Offset Addr § 95 (5Fh). See 2.3 "Byte-Orie I Instructions et Mode" for d	essing mode ented and in Indexed			set is enable Indexed Lite whenever f ≤ Section 23.2 Bit-Oriented	d the extended d, this instruction ral Offset Addro 3 95 (5Fh). See 2.3 "Byte-Orie I Instructions et Mode" for de	on operates in essing mode nted and in Indexed	
Words:	1			Words	5:	1			
Cycles:	•	cles if skip and 2-word instruc		Cycles	S:		/cles if skip and a 2-word instru		
Q Cycle Activity	/:			Q Cy	cle Activity:				
Q1	Q2	Q3	Q4	_	Q1	Q2	Q3	Q4	
Decode		Process	No		Decode	Read	Process	No	
	register 'f'	Data	operation	16 - 1 -		register 'f'	Data	operation	
If skip:	02	02	04	lf skip		00	02	04	
Q1 No	Q2 No	Q3 No	Q4 No	Г	Q1 No	Q2 No	Q3 No	Q4 No	
operation	-	operation	operation		operation	operation	operation	operation	
If skip and follo	wed by 2-word ins	truction:		lf skip	o and followed	d by 2-word inst	truction:		
Q1	Q2	Q3	Q4	_	Q1	Q2	Q3	Q4	
No	No	No	No		No	No	No	No	
operation	n operation	operation	operation	_	operation	operation	operation	operation	
No operation	No operation	No operation	No operation		No operation	No operation	No operation	No operation	
<u>Example:</u>	HERE BI FALSE : TRUE :	IFSC FLAG	5, 1, O	<u>Exam</u>	<u>ple:</u>	HERE BI FALSE : TRUE :	TFSS FLAG	, 1, 0	
If FLA	= add ction G<1> = 0; PC = add G<1> = 1;	ress (HERE) ress (TRUE) ress (FALSE))		Before Instruc PC After Instructio If FLAG< PC If FLAG< PC	tion = add n 1> = 0; = add 1> = 1;	ress (HERE) ress (False) ress (TRUE)		

BTG		Bit Toggle	f		BOV		Branch if C	Overflow			
Synta	ax:	BTG f, b {,a	ı}		Synta	x:	BOV n				
Oper	ands:	$0 \le f \le 255$			Opera	ands:	-128 ≤ n ≤ 1	27			
		0 ≤ b < 7 a ∈ [0,1]			Opera	ation:		if Overflow bit is '1', (PC) + 2 + 2n \rightarrow PC			
Oper	ation:	$(\overline{f} < b >) \to f <$:b>		Status	s Affected:	None				
Statu	is Affected:	None			Encoding:		1110	0100 nn	nn nnnn		
Enco	oding:	0111	bbba ff	ff ffff	Desc	ription:	If the Overf	ow bit is '1', t	hen the		
Desc	scription: Bit 'b' in data memory location 'f' is inverted.				program wi	l branch.					
		lf 'a' is '1', t	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).		The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be						
			nd the extend				PC + 2 + 2n. This instruction is then a two-cycle instruction.				
			led, this instru	ction operates Addressing	Word	s:	1				
			never f ≤ 95 (5	•	Cycle	s.	1(2)				
		Bit-Oriente	.2.3 "Byte-Or ed Instruction set Mode" for	s in Indexed		cle Activity:	()				
Word	ls:	1			r	Q1	Q2	Q3	Q4		
Cycle	es:	1				Decode	Read literal 'n'	Process Data	Write to PC		
QC	ycle Activity:					No	No	No	No		
	Q1	Q2	Q3	Q4		operation	operation	operation	operation		
	Decode	Read register 'f'	Process Data	Write register 'f'	If No	Jump: Q1	Q2	Q3	Q4		
		regiotor i	Dulu	regiotor]	Decode	Read literal	Process	No		
Exan	nple:	BTG P	ORTC, 4, ()		Decode	'n'	Data	operation		
	Before Instruc PORTC		0101 [75h]		Exam	ple:	HERE	BOV Jump)		
	After Instruction PORTC		0101 [65h]			Before Instruct PC After Instructi If Overflu PC If Overflu PC	= ad on ow = 1; = ad ow = 0;	dress (HERE dress (Jump dress (HERE)		

ΒZ		Branch if Z	lero					
Synta	ax:	BZ n						
Oper	ands:	-128 ≤ n ≤ 1	127					
Oper	ration:	if Zero bit is (PC) + 2 + 2	,					
Statu	is Affected:	None						
Enco	oding:	1110	0000 nni	nn nnnn				
Desc	cription:		If the Zero bit is '1', then the program will branch.					
		added to the incremente instruction,	d to fetch the r the new addre n. This instruct	e PC will have next ess will be				
Word	ds:	1						
Cycle	es:	1(2)						
Q C If Ju	ycle Activity:							
	Q1	Q2	Q3	Q4				
	Decode	Read literal	Process	Write to				
		ʻn'	Data	PC				
	No operation	No operation	No operation	No operation				
If No	o Jump:	operation	operation	operation				
	Q1	Q2	Q3	Q4				
	Decode	Read literal	Process	No				
		'n'	Data	operation				
Exar	nple:	HERE	BZ Jump					
	Before Instruc PC	= ad	dress (HERE))				
	After Instructio							
	If Zero PC	= 1; = ad	dress (Jump))				
	If Zero PC	= 0; = ad	dress (HERE					
	FU	= au	UICOO (HEKE	+ 2)				

Syntax:		~)			
Cymax.	CALL k {				
Operands:	0 ≤ k ≤ 104 s ∈ [0,1]	0 ≤ k ≤ 1048575 s ∈ [0,1]			
Operation:	$\begin{array}{l} (PC) + 4 - \\ k \rightarrow PC < 2 \\ \text{if } s = 1 \\ (W) \rightarrow WS \\ (STATUS) \\ (BSR) \rightarrow E \end{array}$	0:1>; 5, → STATU	JSS,		
Status Affected:	None				
Encoding: 1st word (k<7:0> 2nd word(k<19:8	·	110s k ₁₉ kkk	k ₇ kk kkk		
	(PC+ 4) is If 's' = 1, th registers a respective STATUSS update occ 20-bit valu CALL is a	ne W, STA re also pu shadow r and BSR curs (defa e 'k' is loa	ATUS a ushed i registe S. If 's' uult). Th uded in	and BSR into their rs, WS, ' = 0, no nen, the to PC<2	ł
Words:	2		niotrat		
	_				
	2				
Cycles:					
		Q3	}	Q4	
Cycles: Q Cycle Activity:	:	Q3 Push P stac	C to	Q4 Read lit 'k'<19: Write to	era 8>
Cycles: Q Cycle Activity: Q1	: Q2 Read literal	Push P	C to k	Read lit 'k'<19:	era 8> P
Cycles: Q Cycle Activity: Q1 Decode No operation	: Q2 Read literal 'k'<7:0>, No operation	Push P stac No operat	C to k ion	Read lit 'k'<19: Write to No operat	era 8> P
Cycles: Q Cycle Activity: Q1 Decode No operation Example:	: Q2 Read literal 'k'<7:0>, No operation HERE	Push P stac No	C to k	Read lit 'k'<19: Write to No operat	era 8> P
Cycles: Q Cycle Activity: Q1 Decode No operation	: Q2 Read literal 'k'<7:0>, No operation HERE	Push P stac No operat	C to k ion THER	Read lit 'k'<19: Write to No operat	era 8> P

CLRF	Clear f					
Syntax:	CLRF f{,	CLRF f {,a}				
Operands:	0 ≤ f ≤ 255 a ∈ [0,1]					
Operation:	$\begin{array}{l} 000h \rightarrow f, \\ 1 \rightarrow Z \end{array}$					
Status Affected:	Z					
Encoding:	0110	101a	ffff	ffff		
Description:	Clears the register.	contents	of the spe	ecified		
	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).					
	set is enabl in Indexed mode wher Section 23 Bit-Oriente	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 23.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.				
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3		Q4		
Decode	Read	Proce		Write		
	register 'f'	Data	a re	gister 'f'		
Example: Before Instruc FLAG_R	EG = 5A	_	REG,1			
After Instructio FLAG_R		h				

CLRWDT	Clear Wate	chdog Ti	mer			
Syntax:	CLRWDT	CLRWDT				
Operands:	None	None				
Operation:						
Status Affected: TO, PD						
Encoding:	0000	0000	0000	0100		
Description: CLRWDT instruction resets the Watchdog Timer. It also resets the postscaler of the WDT. Status bits, and PD, are set.				ts the		
Words:	1	1				
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3	1	Q4		
Decode	No operation	Proce Data		No peration		
Example: CLRWDT Before Instruction WDT Counter = ?						

00h

= 0 = 1 = 1

=

After Instruction WDT Counter

WDT Postscaler TO PD

COMF	Compleme	ent f		CPF	SEQ	Compare f	with W, Skip	if f = W
Syntax:	COMF f	{,d {,a}}		Synt	ax:	CPFSEQ	f {,a}	
Operands:	0 ≤ f ≤ 255			Oper	rands:	$0 \leq f \leq 255$		
	d ∈ [0,1]					a ∈ [0,1]		
	a ∈ [0,1]			Oper	ration:	(f) – (W),		
Operation:	$\overline{f} \rightarrow dest$					skip if (f) = (unsigned c	(W) comparison)	
Status Affected:	N, Z		i	Statu	s Affected:	None		
Encoding:	0001	11da ff	ff ffff	Enco	oding:	0110	001a ff:	ff ffff
Description:	complemer stored in W	nts of register ' nted. If 'd' is '0 /. If 'd' is '1', th < in register 'f'	', the result is e result is	Desc	cription:	location 'f' t	the contents of the contents an unsigned s	
	lf 'a' is 'o', t	he Access Ba he BSR is use	nk is selected. Ind to select the			discarded a	en the fetchec and a NOP is e aking this a two	xecuted
	set is enab in Indexed	and the extend led, this instru- Literal Offset i never f \leq 95 (5	ction operates Addressing				he BSR is use	nk is selected. d to select the
	Section 23 Bit-Oriente	ad Instruction set Mode" for	iented and is in Indexed			set is enabl in Indexed	nd the extended ed, this instruct Literal Offset A bever f \leq 95 (5)	ction operates Addressing
Words:	1						.2.3 "Byte-Or	,
Cycles:	1						d Instruction	
Q Cycle Activity:						Literal Offs	set Mode" for	details.
Q1	Q2	Q3	Q4	Word	ds:	1		
Decode	Read register 'f'	Process Data	Write to destination	Cycle	es:	•	cles if skip and 2-word instrue	
				0.0	ycle Activity:	,		
Example:	COMF	REG, 0, 0			Q1	Q2	Q3	Q4
Before Instruct REG	ion = 13h				Decode	Read register 'f'	Process Data	No operation
After Instruction				lf sk	kip:			
REG W	= 13h = ECh				Q1	Q2	Q3	Q4
					No	No	No	No
					operation	operation	operation	operation
				lt sk	•	d by 2-word in	-	04
					Q1 No	Q2	Q3 No	Q4
					operation	No operation	operation	No operation
					No	No	No	No
					operation	operation	operation	operation
				<u>Exar</u>	nple:	HERE NEQUAL EQUAL	CPFSEQ REG : :	3, 0
					Before Instruc		22	

PC Address = W = REG =

After Instruction If REG PC If REG PC here ? ?

= = ≠ W; Address (EQUAL) W; Address (NEQUAL)

CPFS	GT	Compare f	with W, Skip	if f > W				
Synta	ax:	CPFSGT	f {,a}					
Opera		0 ≤ f ≤ 255						
		a ∈ [0,1]						
Opera	ation:	(f) - (W),						
		skip if (f) > ((W)					
		(unsigned c	omparison)					
Statu	s Affected:	None						
Enco	ding:	0110	010a fff	f ffff				
Desc	ription:	location 'f' to	he contents of o the contents an unsigned s	of the W by				
		contents of instruction is	nts of 'f' are gro WREG, then t s discarded ar stead, making struction.	he fetched nd a NOP is				
			he Access Bar he BSR is used (default).					
		set is enabl in Indexed I mode when Section 23 Bit-Oriente	nd the extende ed, this instruct Literal Offset A ever $f \le 95$ (5F .2.3 "Byte-Ori d Instructions set Mode" for	ation operates addressing Fh). See ented and s in Indexed				
Word	s:	1						
Cycle	es:		ycles if skip ar a 2-word instru					
QC	cle Activity:							
r	Q1	Q2	Q3	Q4				
	Decode	Read	Process	No				
lf ski		register 'f'	Data	operation				
11 54	ιμ. Q1	Q2	Q3	04				
Ī	No	No	No	No				
	operation	operation	operation	operation				
lf ski	ip and followed	d by 2-word ins						
	Q1	Q2	Q3	Q4				
	No	No	No	No				
	operation	operation	operation	operation				
	No	No	No	No				
l	operation	operation	operation	operation				
<u>Exarr</u>	<u>iple:</u>	HERE NGREATER GREATER	CPFSGT RE : :	G, 0				
I	Before Instruc	tion						
	PC W		dress (HERE))				
	vv After Instructio	•						
4	If REG	> W;						
	PC	= Ad	dress (GREAT	FER)				
	If REG PC	≤ W; = Ad	dress (NGRE2	ATER)				

CPF	SLT	Compare f with W, Skip if f < W				
Synta	ax:	CPFSLT	CPFSLT f {,a}			
Oper	rands:	$0 \le f \le 255$	$0 \le f \le 255$			
		a ∈ [0,1]				
Oper	ration:	(f) - (W),				
		skip if (f) <	· · /			
0			comparison)			
_	is Affected:	None				
Enco	oding:	0110		fff ffff		
Desc	pription:	location 'f'	the contents to the conten an unsigned	•		
		If the contents of 'f' are less than the contents of W, then the fetched instruction is discarded and a NOP is executed instead, making this a				
		two-cycle in				
		,	he BSR is us	ank is selected. ed to select the		
Word	ds:	1	· ,			
Cycle	es:	1(2)				
e jen		()	ycles if skip a	and followed		
		by	a 2-word inst	ruction.		
QC	ycle Activity:					
	Q1	Q2	Q3	Q4		
	Decode	Read	Process Data	No		
lf sk	rin:	register 'f'	Dala	operation		
11 51	ωρ. Q1	Q2	Q3	Q4		
	No	No	No	No		
	operation	operation	operation	operation		
lf sk	ip and followe	d by 2-word in	struction:			
	Q1	Q2	Q3	Q4		
	No	No	No	No		
	operation No	operation No	operation No	operation No		
	operation	operation	operation	operation		
<u>Exan</u>	<u>nple:</u>	NLESS	CPFSLT REC : :	3, 1		
	Before Instruc	tion				
	PC		dress (HER	E)		
	W After Instructio	= ?				
	If REG	< W				
	PC If REG	= Ac ≥ W	ldress (LES	S)		
	PC		, dress (NLE	SS)		

DAW	Decimal A	djust W Regis	ster	DECF	Decremen	tf	
Syntax:	DAW			Syntax:	DECF f{,	d {,a}}	
Operands:	None			Operands:	$0 \le f \le 255$		
Operation:		> 9] or [DC = 1			d ∈ [0,1] a ∈ [0,1]		
	(W<3:0>) + else	$6 \rightarrow W < 3:0>;$		Operation:	$a \in [0, 1]$ (f) $-1 \rightarrow de$	oct .	
	(W<3:0>) –	→ W<3:0>		Status Affected:	C, DC, N, C		
		01	lle e e	Encoding:	0000		ff ffff
	•	> 9] or [C = 1], $6 \rightarrow W < 7:4>;$		Description:		register 'f'. If	
	C = 1;	• / 11 11 10 ,		Description.		red in W. If 'c	
	else	N/ 27:45				red back in re	egister 'f'
Status Affected:	(W<7:4>) – C	→ VV<7.4>			(default).		
	-		0.0111			he Access Ba he BSR is us	
Encoding:	0000	0000 000			GPR bank		
Description:		ts the eight-bit om the earlier a			lf 'a' is '0' a	nd the extend	led instructio
	variables (e	each in packed	BCD format)		set is enabled, this instruction		
	and produc result.	es a correct pa	acked BCD			Literal Offset iever f ≤ 95 (5	•
Words:	1				Section 23	.2.3 "Byte-O	riented and
Cycles:	1					ed Instruction set Mode" for	
Q Cycle Activity:				Words:	1		detailo.
Q Oycle Activity. Q1	Q2	Q3	Q4	Cycles:	1		
Decode	Read	Process	Write	Q Cycle Activity:			
	register W	Data	W	Q1	Q2	Q3	Q4
Example 1:	DAW			Decode	Read	Process	Write to
Before Instru					register 'f'	Data	destination
W	= A5h						
C DC	= 0 = 0			Example:		CNT, 1, ()
After Instruc				Before Instru CNT	iction = 01h		
W C	= 05h = 1			Z	= 0		
DC	= 0			After Instruct CNT			
Example 2:				Z	= 00h = 1		
Before Instru							
W C	= CEh = 0						
DC	= 0						
After Instruc							
۱۸/	-						
W C DC	= 1						

DEC	FSZ	Decrement	f, Skip if 0			
Synta	ax:	DECFSZ f {,d {,a}}				
Oper	ands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$	$0 \le f \le 255$ d \equiv [0,1]			
Oper	ration:	(f) – 1 \rightarrow de skip if result	-			
Statu	is Affected:	None				
Enco	oding:	0010	11da fff	f ffff		
Desc	cription:	decremente placed in W	ts of register 'f ed. If 'd' is '0', f /. If 'd' is '1', th < in register 'f'	the result is le result is		
		which is alro and a NOP i	is '0', the next eady fetched in s executed inst le instruction.	s discarded		
			ne Access Bar ne BSR is use (default).			
		set is enabl in Indexed I mode when Section 23 Bit-Oriente	nd the extended ed, this instruct Literal Offset A lever $f \le 95$ (5) .2.3 "Byte-Or ad Instruction set Mode" for	ction operates Addressing Fh). See iented and s in Indexed		
Word	ds:	1				
Cycle	es:		rcles if skip an 2-word instru			
QC	ycle Activity:					
	Q1	Q2	Q3	Q4		
	Decode	Read	Process	Write to		
lf sk	in:	register 'f'	Data	destination		
11 51	ωρ. Q1	Q2	Q3	Q4		
	No	No	No	No		
	operation	operation	operation	operation		
lf sk	ip and followe	d by 2-word in	struction:			
	Q1					
		Q2	Q3	Q4		
	No	No	No	No		
	No operation	No operation	No operation	No operation		
	No	No	No	No		
Exan	No operation No	No operation No	No operation No	No operation No		
Exan	No operation No operation	No operation No operation HERE CONTINUE	No operation No operation	No operation No operation		
<u>Exan</u>	No operation No operation nple: Before Instruct PC After Instruction	No operation No operation HERE CONTINUE stion = Address	No operation No operation DECFSZ GOTO S (HERE)	No operation No operation		
Exan	No operation No operation nple: Before Instruc PC	No operation No operation HERE CONTINUE tion = Address on = CNT - 1 = 0;	No operation No operation DECFSZ GOTO S (HERE)	No operation No operation		
Exan	No operation No operation nple: Before Instruct PC After Instructio CNT If CNT	No operation No operation HERE CONTINUE tion = Address on = CNT - 1 = 0;	No operation No operation DECFSZ GOTO S (HERE)	No operation No operation CNT, 1, 1 LOOP		

DCFS	DCFSNZ Decrement f, Skip if Not 0					
Synta	x:	DCFSNZ	f {,d {,a}}			
$\begin{array}{llllllllllllllllllllllllllllllllllll$						
Operation: $(f) - 1 \rightarrow dest,$ skip if result $\neq 0$						
Status	s Affected:	None				
Enco	ding:	0100 11da ffff ffff				
Desci	ription:	decremente placed in V	ts of register 'f ed. If 'd' is 'o', t /. If 'd' is '1', th k in register 'f'	the result is le result is		
		instruction discarded a	is not '0', the which is alread and a NOP is ex aking it a two-c	ly fetched is kecuted		
			he Access Bar he BSR is usec (default).			
If 'a' is '0' and the extended instruct set is enabled, this instruction opera in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 23.2.3 "Byte-Oriented an Bit-Oriented Instructions in Index Literal Offset Mode" for details.						
Word	s:	1				
Cycle	S:		ycles if skip ar a 2-word instru			
QCy	cle Activity:					
-	Q1	Q2	Q3	Q4		
	Decode	Read	Process	Write to		
ال مان		register 'f'	Data	destination		
lf ski	ρ: Q1	Q2	Q3	Q4		
Г	No	No	No	No		
	operation	operation	operation	operation		
lf ski	p and followe	d by 2-word in	struction:			
г	Q1	Q2	Q3	Q4		
	No	No	No	No		
F	operation No	operation No	operation No	operation No		
	operation	operation	operation	operation		
Exam	•	HERE I		IP, 1, 0		
	Before Instruc TEMP After Instructic	=	?			
/	TEMP	n =	TEMP – 1,			
	If TEMP	=	0;			
	PC If TEMP	= ≠	0;	ZERO)		
	PC	=	Address (1	NZERO)		

GOT	ю	Unconditional Branch					
Synta	ax:	GOTO k					
Oper	rands:	$0 \le k \le 104$	0 ≤ k ≤ 1048575				
Oper	ration:	$k \rightarrow PC<2$	$k \rightarrow PC < 20:1 >$				
Statu	is Affected:	None					
1st w	oding: vord (k<7:0>) word(k<19:8>)	1110 1111	1111 k ₁₉ kkk	k ₇ kk kkkł	0		
Desc	cription:	range. The	within ent 20-bit va GOTO is	re 2-MI lue 'k' i	nal branch byte memory s loaded into a two-cycle		
Word	ds:	2					
Cycle	es:	2					
QC	ycle Activity:						
	Q1	Q2	Q3		Q4		
	Decode	Read literal 'k'<7:0>,	No operat	ion	Read literal 'k'<19:8>, Write to PC		
	No	No	No		No		
	operation	operation	operat	ion	operation		
<u>Exan</u>	nple: After Instructic PC =	GOTO THE on Address (1					

INCF	Increment	f				
Syntax:	INCF f{,c	l {,a}}				
Operands:	0 ≤ f ≤ 255 d ∈ [0,1] a ∈ [0,1]	0 ≤ f ≤ 255 d ∈ [0,1]				
Operation:	(f) + 1 \rightarrow de	est				
Status Affected:	C, DC, N,	OV, Z				
Encoding:	0010	10da f	fff	ffff		
Description:	incremente placed in W	ts of registe d. If 'd' is 'o' /. If 'd' is '1', k in register	, the re the re	esult is sult is		
	,	he Access E he BSR is u (default).				
	in Indexed mode wher Section 23 Bit-Oriente	ed, this inst Literal Offse never f ≤ 95 .2.3 "Byte- ed Instructionset Mode" f	t Addre (5Fh). Oriente ons in	essing See ed and Indexed		
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3		Q4		
Decode	Read	Process		Vrite to		
	register 'f'	Data	de	stination		
Example:	INCF	CNT, 1,	0			
Before Instruc CNT Z C	tion = FFh = 0 = ?					

INCF	SZ	Increment	f, Skip if	0	
Syntax: INCFSZ f {,d {,a}}					
Oper	$\begin{array}{llllllllllllllllllllllllllllllllllll$				
Operation: (f) + 1 \rightarrow dest, skip if result = 0					
Statu	s Affected:	None			
Enco	ding:	0011	11da	ffff	ffff
Desc	ription:	The content incremented placed in W placed back	d. If 'd' is ⁄. If 'd' is	'0', the r '1', the re	esult is esult is
		If the result which is alre and a NOP i it a two-cycl	eady feto s execut	hed is di ed instea	scarded
		lf 'a' is 'o', th If 'a' is '1', th GPR bank (ne BSR i		
		If 'a' is '0' ar set is enabl in Indexed I mode when Section 23. Bit-Oriente Literal Offs	ed, this i ∟iteral Of ever f ≤ : .2.3 "By† d Instru	nstructior fset Addr 95 (5Fh). te-Orient ctions in	n operates ressing See red and Indexed
Word	s:	1			
Cycle	es:		-	skip and f I instructi	
QC	ycle Activity:				
i	Q1	Q2	Q3		Q4
	Decode	Read register 'f'	Proce Data		Write to estination
lf sk	ip:				
i	Q1	Q2	Q3		Q4
	No	No	No	ion	No
lfek	operation	operation d by 2-word ins	operat		peration
11 51	Q1	Q2	Q3		Q4
	No	No	No		No
	operation	operation	operat	ion o	peration
	No	No	No	.	No
	operation	operation	operat	ion o	peration
<u>Exan</u>	nple:	HERE 1 NZERO : ZERO :		CNT,	1, 0
	Before Instruc PC	= Address	G (HERE)	
	After Instructic CNT If CNT	on = CNT + 1 = 0;	I		
	PC	= Address	(ZERO)	
	If CNT PC	≠ 0; = Address			

INFSNZ Increment f, Skip if Not 0							
Synta	ax:	INFSNZ f	{,d {,a}}				
Oper	ands:	0 ≤ f ≤ 255					
		$d\in [0,1]$					
		a ∈ [0,1]					
Oper	ation:	(f) + 1 \rightarrow de					
		skip if resul	t ≠ 0				
Statu	is Affected:	None					
Enco	oding:	0100	10da ffi				
Desc	ription:	incremente placed in W	The contents of register 'f' are incremented. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default).				
If the result is not 'o', the next instruction which is already fer discarded and a NOP is execu instead, making it a two-cycle instruction.				ly fetched is cecuted			
			he Access Bar he BSR is use (default).				
	If 'a' is '0' and the extended instruction set is enabled, this instruction operate in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 23.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.						
Word	ls:	1					
Cycle		1(2)					
Oyok		· · /	cles if skip an	d followed			
		by a	a 2-word instru	ction.			
QC	ycle Activity:						
	Q1	Q2	Q3	Q4			
	Decode	Read	Process	Write to			
		register 'f'	Data	destination			
lf sk	•						
	Q1	Q2	Q3	Q4			
	No operation	No	No	No			
		oporation	oporation				
lf ck		operation	operation	operation			
lf sk	ip and followe	d by 2-word in	struction:	operation			
lf sk	ip and followe Q1	d by 2-word in: Q2	struction: Q3	operation Q4			
lf sk	ip and followe Q1 No	d by 2-word in: Q2 No	struction: Q3 No	operation Q4 No			
lf sk	ip and follower Q1 No operation	d by 2-word in: Q2 No operation	struction: Q3 No operation	operation Q4 No operation			
lf sk	ip and followe Q1 No	d by 2-word in: Q2 No	struction: Q3 No	operation Q4 No			
lf sk <u>Exan</u>	Q1 No operation No operation	d by 2-word in: Q2 No operation No operation	struction: Q3 No operation No	Q4 Q4 No operation No operation			
Exan	ip and follower Q1 No operation No operation nple: Before Instruc PC	d by 2-word in: Q2 No operation No operation HERE ZERO NZERO tion = Address	struction: Q3 No operation No operation	Q4 Q4 No operation No operation			
Exan	ip and follower Q1 No operation No operation nple: Before Instruc	d by 2-word in: Q2 No operation No operation HERE ZERO NZERO tion = Address	struction: Q3 No operation No operation INFSNZ REG (HERE)	Q4 Q4 No operation No operation			

IORI	_w	Inclusive OR Literal with W					
Synt	ax:	IORLW k					
Ope	rands:	$0 \le k \le 25$	5				
Ope	ration:	(W) .OR. k	$x \to W$				
Status Affected: N, Z							
Encoding: 0000 1001 kkkk kkk					kkkk		
Desc	cription:	The conte eight-bit lit in W.					
Word	ds:	1	1				
Cycl	es:	1	1				
QC	ycle Activity:						
	Q1	Q2	Q3	3	Q4		
	Decode	Read literal 'k'	Proce Data		Write to W		
<u>Exar</u>	<u>nple:</u>	IORLW	35h				
Before Instruction W = 9Ah After Instruction							

BFh

=

W

IORWF	Inclusive (OR W wit	h f			
Syntax:	IORWF	f {,d {,a}}				
Operands:	0 ≤ f ≤ 255 d ∈ [0,1] a ∈ [0,1]	d ∈ [0,1] a ∈ [0,1]				
Operation:	(W) .OR. (f	(W) .OR. (f) \rightarrow dest				
Status Affected:	N, Z					
Encoding:	0001	00da	ffff	ffff		
Description:	Inclusive OR W with register 'f'. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default).					
	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).					
	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 23.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.					
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3		Q4		
Decode	Read register 'f'	Proce Data		Write to destination		
Example: Before Instruc	tion	ESULT,	0, 1			

Before Instruction	

RESULT	=	13h
W	=	91h
After Instructio	n	
RESULT	=	13h
W	=	93h

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LFSF	2	Load FSR					
Synta	ax:	LFSR f, k	LFSR f, k				
Operands:		$0 \le f \le 2$ $0 \le k \le 409$	$\begin{array}{l} 0 \leq f \leq 2 \\ 0 \leq k \leq 4095 \end{array}$				
Oper	ation:	$k\toFSRf$					
Statu	s Affected:	None					
Enco	ding:	1110 1111	1110 0000	00ff k ₇ kkk	k ₁₁ kkk kkkk		
Description: The 12-bit literal 'k' is loaded into the file select register pointed to by 'f'.							
Word	ls:	2					
Cycles:		2	2				
QC	ycle Activity:						
	Q1	Q2	Q3		Q4		
	Decode	Read literal 'k' MSB	Proce: Data	ι lit Ν	Write eral 'k' ISB to SRfH		
	Decode	Read literal 'k' LSB	Proce: Data		te literal o FSRfL		
<u>Exan</u>	<u>nple:</u>	LFSR 2,	3ABh				
	After Instruction FSR2H = 03h FSR2L = ABh						

MOVF	Move f					
Syntax:	MOVF f{	,d {,a}}				
Operands:	0 ≤ f ≤ 255 d ∈ [0,1] a ∈ [0,1]	d ∈ [0,1]				
Operation:	$f \to \text{dest}$					
Status Affected:	N, Z	N, Z				
Encoding:	0101	00da fi	fff ffff			
Description:	a destinatio status of 'd' placed in W placed bacl	The contents of register 'f' are moved to a destination dependent upon the status of 'd'. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default). Location 'f' can be anywhere in the 256-byte bank				
		he BSR is us	ank is selected. ed to select the			
	set is enabl in Indexed mode wher Section 23 Bit-Oriente	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \le 95$ (5Fh). See Section 23.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.				
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3	Q4			
Decode	Read register 'f'	Process Data	Write W			
<u>Example:</u> Before Instruc REG		EG, 0, 0 h				
After Instruction						

22h 22h

=

REG W

MOVFF	Move f to f				
Syntax:	MOVFF f _s	,f _d			
Operands:	$\begin{array}{l} 0 \leq f_{s} \leq 409 \\ 0 \leq f_{d} \leq 409 \end{array}$				
Operation:	$(f_s) \to f_d$				
Status Affected:	None				
Encoding: 1st word (source) 2nd word (destin.)	1100 1111	ffff ffff	ffff ffff	ffff _s ffff _d	
Description:	The conten moved to d Location of in the 4096 FFFh) and can also be FFFh.	estination source 'f -byte dat location	n register s' can be a space (lo of destina	'f _d '. anywhere 000h to tion 'f _d '	
	Either source or destination can be W (a useful special situation).				
	MOVFF is particularly useful for transferring a data memory location to a peripheral register (such as the transmi buffer or an I/O port).				
	The MOVFF PCL, TOSU destination	J, TOSH			
Words:	2				
Cycles:	2				
Q Cycle Activity:					
Q1	Q2	Q3	l	Q4	
Decode	Read register 'f' (src)	Proce Data		No peration	
Decode	No operation No dummy	No operat	ion re	Write gister 'f' (dest)	

Before Instruction		
REG1	=	33h
REG2	=	11h
After Instruction		
REG1	=	33h
REG2	=	33h

MOVLB	Move Literal to Low Nibble in BSR					
Syntax:	MOVLW I	<				
Operands:	$0 \le k \le 255$	$0 \le k \le 255$				
Operation:	$k \rightarrow BSR$					
Status Affected:	None					
Encoding:	0000	0001	kkk	:k	kkkk	
Description:	The eight-bit literal 'k' is loaded into the Bank Select Register (BSR). The value of BSR<7:4> always remains '0' regardless of the value of k ₇ :k ₄ .					
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3	}		Q4	
Decode	Read literal 'k'	Proce Data		••••	te literal to BSR	
Example:	MOVLB	5				
Before Instruction BSR Register = 02h After Instruction						

05h

BSR Register =

Move W to f

MOVWF

MOVLW Move Literal to W							
Synta	ax:	MOVLW	k				
Oper	ands:	$0 \le k \le 25$	5				
Oper	ation:	$k\toW$					
Statu	s Affected:	None					
Encoding: 0000 1110 kkkk kk				kkkk			
Desc	ription:	The eight-	The eight-bit literal 'k' is loaded into W.				
Word	ls:	1					
Cycle	es:	1					
QC	ycle Activity:						
	Q1	Q2	Q3	;	Q4		
	Decode	Read	Proce	SS	Write to		
		literal 'k'	Data	a	W		
Exan	nple:	MOVLW	5Ah				

After Instruction W

=

5Ah

Syntax:	MOVWF	f {,a}				
Operands:	0 ≤ f ≤ 255 a ∈ [0,1]					
Operation:	$(W) \rightarrow f$					
Status Affected:	None					
Encoding:	0110	111a	ffff	ffff		
Description:	Location 'f'	Move data from W to register 'f'. Location 'f' can be anywhere in the 256-byte bank.				
	lf 'a' is '1', t	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).				
	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 23.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.					
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3		Q4		
Decode	Read	Proces	ss	Write		
	register 'f'	Data	re	gister 'f'		
Example:	MOVWF	REG, 0				
Before Instruct	tion					
W REG	= 4Fh = FFh					
REG = FFh						

=

4Fh 4Fh

After Instruction

W REG

MULLW	Multiply Li	Multiply Literal with W						
Syntax:	MULLW	k						
Operands:	$0 \le k \le 255$	$0 \le k \le 255$						
Operation:	(W) x k \rightarrow	(W) x k \rightarrow PRODH:PRODL						
Status Affected:	None	None						
Encoding:	0000	1101 kkl	kk kkkk					
Description:	out betwee 8-bit literal placed in th	An unsigned multiplication is carried out between the contents of W and the 8-bit literal 'k'. The 16-bit result is placed in the PRODH:PRODL register pair. PRODH contains the high byte.						
	W is uncha	W is unchanged.						
	None of the	None of the Status flags are affected.						
	possible in	Note that neither Overflow nor Carry is possible in this operation. A Zero result is possible but not detected.						
Words:	1	1						
Cycles:	1	1						
Q Cycle Activity:								
Q1	Q2	Q3	Q4					
Decode	Read literal 'k'	Process Data	Write registers PRODH: PRODL					
Example:	MULLW	0C4h						
Before Instruc	tion							
W PRODH	= E2 = ? = ?	?h						
PRODL After Instruction	•							

MULWF	I	Multiply W with f						
Syntax:	l	MULWF f {,a}						
Operands:		0 ≤ f ≤ 255 a ∈ [0,1]						
Operation:	((W) x (f) \rightarrow PRODH:PRODL						
Status Affe	cted: I	None						
Encoding:	[0000	0)1a	ffff	-	ffff	
Description:		An unsigned multiplication is carried out between the contents of W and the register file location 'f'. The 16-bit result is stored in the PRODH:PRODL register pair. PRODH contains the high byte. Both W and 'f' are unchanged.						
	I	None of the Status flags are affected.						
	I	Note that neither Overflow nor Carry is possible in this operation. A Zero result is possible but not detected.						
		If 'a' is '0', the Access Bank is selected. I 'a' is '1', the BSR is used to select the GPR bank (default).						
	i	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \le 95$ (5Fh). See Section 23.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.						
Words:		1						
Cycles:		1						
Q Cycle A	ctivity:							
	Q1	Q2		C	13		Q4	
Dec	code	Read register 'l	p	Proc Da		re Pl	Write gisters RODH: RODL	
Evampla		MTTT 1.7-		DEC	1			
Example: Before	e Instruct	MULWF		REG,	Ŧ			
V F F After I V	V REG PRODH PRODL nstructio V	= = = n =	C4ł B5ł ? C4ł	า				
	RODH	=	B5ł 8Ał					

NEGF	Negate f			
Syntax:	NEGF f {,a}			
Operands:	0 ≤ f ≤ 255 a ∈ [0,1]			
Operation:	$(\overline{f}) + 1 \rightarrow f$			
Status Affected:	N, OV, C, DC, Z			
Encoding:	0110 110a ffff ffff			
Description:	Location 'f' is negated using two's complement. The result is placed in the data memory location 'f'.			
	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).			
	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \le 95$ (5Fh). See Section 23.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.			
Words:	1			
Cycles:	1			
O Cycle Activity:				

NOP	NOP No Operation						
Synta	ax:	NOP	NOP				
Oper	ands:	None					
Oper	ation:	No operation					
Status Affected: None							
Encoding:		0000 1111	0000 xxxx	000 xxx	-	0000 xxxx	
Desc	ription:	No operati	on.				
Word	ls:	1					
Cycle	es:	1					
QC	ycle Activity:						
	Q1	Q2	Q	3		Q4	
	Decode	No operation	No operat		ор	No eration	

Example:

None.

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read	Process	Write
	register 'f'	Data	register 'f'

Example: NEGF REG, 1

> Before Instruction REG = 0011 1010 [3Ah] After Instruction REG = 1100 0110 [C6h]

РОР	Рор Тор о	Pop Top of Return Stack				
Syntax:	POP					
Operands:	None					
Operation:	(TOS) ightarrow b	it bucket				
Status Affected:	None					
Encoding:	0000	0000	000	0 0110		
Description:	stack and i then becor was pushe This instru the user to	The TOS value is pulled off the return stack and is discarded. The TOS value then becomes the previous value that was pushed onto the return stack. This instruction is provided to enable the user to properly manage the return stack to incorporate a software stack.				
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3		Q4		
Decode	No operation	POP T valu		No operation		
Example:	POP GOTO	NEW				
Before Instru TOS Stack (*	ction level down)		031A2 14332			
After Instruct TOS PC	ion		14332 IEW	h		

PUS	н	Push Top o	Push Top of Return Stack			
Synt	ax:	PUSH				
Oper	rands:	None	None			
Oper	ration:	$({\sf PC}$ + 2) $ ightarrow$	тоѕ			
Statu	us Affected:	None				
Enco	oding:	0000	0000 0000 0000			0101
Desc	cription:	The PC + 2 the return s value is pus This instruc software sta then pushin	tack. shed c tion a ack by	The prev lown on llows imp modifyii	ious the s blem ng Tr	TOS stack. enting a OS and
Word	ds:	1	1			
Cycle	es:	1				
QC	ycle Activity:					
	Q1	Q2		Q3		Q4
	Decode	PUSH PC + 2 onto return stack	No operation		ор	No peration
Exar	<u>nple:</u>	PUSH				
<u>Exar</u>	<u>nple:</u> Before Instruc TOS PC		= =	345Ah 0124h		

RCA	LL	Relative Ca	all				
Synta	ax:	RCALL n	RCALL n				
Oper	ands:	-1024 ≤ n ≤	1023				
Oper	ation:	· · /	$\begin{array}{l} (PC) + 2 \rightarrow TOS, \\ (PC) + 2 + 2n \rightarrow PC \end{array}$				
Statu	Status Affected: None						
Enco	ding:	1101	1101 1nnn nnnn nnnn				
Desc	ription:	Subroutine call with a jump up to 1K from the current location. First, return address (PC + 2) is pushed onto the stack. Then, add the 2's complement number '2n' to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is a two-cycle instruction.					
Word	ls:	1	1				
Cycle	es:	2					
QC	ycle Activity:						
	Q1	Q2	Q3	3		Q4	
	Decode	Read literal 'n' PUSH PC to stack	Proce Data		Writ	e to PC	
	No	No	No			No	
	operation	operation	operat	lion	ope	eration	

Example:	HERE	RCALL Jump

Before Instruction

PC = Address (HERE) After Instruction PC = Address (Jump) TOS = Address (HERE + 2)

RES	ET	Reset					
Synta	ax:	RESET	RESET				
Oper	ands:	None	None				
Oper	ation:	Reset all registers and flags that are affected by a $\overline{\text{MCLR}}$ Reset.					
Status Affected: All							
Enco	ding:	0000	0000	0000 1111			
Description: This instruction provides a way to execute a MCLR Reset in softwar							
Word	ls:	1	1				
Cycle	es:	1					
QC	ycle Activity:						
	Q1	Q2	Q3	1	Q4		
	Decode	Start	No		No		
		reset	operat	ion op	eration		

Example:

After Instruction

Registers = Flags* =	Reset Value Reset Value
riago =	rieser value

RESET

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RET	FIE	Return from	m Interrupt		RET	LW	Return Lite	eral to W	
Synt	ax:	RETFIE {	s}		Synt	ax:	RETLW k		
Ope	rands:	$s \in [0,1]$			Oper	rands:	$0 \le k \le 255$		
Ope	ration:	$(TOS) \rightarrow P$ 1 \rightarrow GIE/G if s = 1	C, IEH or PEIE/G	àIEL;	Oper	ration:	$k \rightarrow W$, (TOS) $\rightarrow P$ PCLATU, P	C, CLATH are u	nchanged
		$(WS) \rightarrow W,$			Statu	us Affected:	None		
		(STATUSS) (BSRS) \rightarrow	\rightarrow STATUS, BSR,		Enco	oding:	0000	1100 kk	kk kkkk
		· · ·	CLATH are ur	nchanged	Desc	cription:	W is loaded	d with the eigh	nt-bit literal 'k'.
Statu	us Affected:	GIE/GIEH,	PEIE/GIEL.						oaded from the
	oding:	0000	0000 00				The high ac	tack (the retu dress latch (,
Desc	cription:		n interrupt. Sta Stack (TOS) is		14/-		remains un	cnangeo.	
		•	errupts are ena		Word		1		
		•	er the high or		Cycl		2		
		0	the shadow re	t. If 's' = 1, the eqisters WS.	QC	Sycle Activity:	00	0.0	<i></i>
			and BSRS are	•		Q1	Q2	Q3	Q4
		STATUS ar	ponding regist nd BSR. If 's' =	0, no update		Decode	Read literal 'k'	Process Data	POP PC from stack, write to W
		of these reo	gisters occurs	(default).		No	No	No	No
Word	ds:	1				operation	operation	operation	operation
Cycl	es:	2							
QC	cycle Activity:				Exar	<u>mple:</u>			
	Q1	Q2	Q3	Q4			W. seebe	lug tabla	
	Decode	No operation	No operation	POP PC from stack		CALL IABLE	; W conta: ; offset v		
		operation	operation	Set GIEH or			; W now ha		
				GIEL			; table va	alue	
	No	No	No	No	TABI	: 1.E			
	operation	operation	operation	operation	11101	ADDWF PCL	; W = off:	set	
						RETLW k0	; Begin ta	able	
<u>Exar</u>	<u>mple:</u>	RETFIE	1			RETLW k1	;		
	After Interrupt	t				:			
	PC W BSR		= TOS = WS = BSRS			RETLW kn	; End of t	table	
	STATUS	H, PEIE/GIEL	= STATU = 1	JSS		Before Instruc	ction		
			- 1			W	= 07h		
						After Instruction	on = value of	fkn	
						vv	- value 0		

RET	URN	Return fro	Return from Subroutine			
Synta	ax:	RETURN	{s}			
Oper	ands:	$s\in [0,1]$				
Oper	ation:	$\begin{array}{l} (\text{TOS}) \rightarrow \text{PC};\\ \text{if } s = 1\\ (\text{WS}) \rightarrow \text{W},\\ (\text{STATUSS}) \rightarrow \text{STATUS},\\ (\text{BSRS}) \rightarrow \text{BSR},\\ \text{PCLATU, PCLATH are unchanged} \end{array}$				
Statu	s Affected:	None				
Enco	ding:	0000	0000	0001	001s	
Description: Return from subroutine. The stack is popped and the top of the stack (TC is loaded into the program counter. 's'= 1, the contents of the shadow registers WS, STATUSS and BSRS loaded into their corresponding registers W, STATUS and BSR. If 's' = 0, no update of these registers occurs (default).				tack (TOS) counter. If nadow d BSRS are ding dSR. If		
Word	ls:	1				
Cycle	es:	2				
QC	ycle Activity:					
	Q1	Q2	Q	3	Q4	
	Decode	No operation	Proce Data		POP PC from stack	
	No	No	No		No	
	operation	operation	operat	ion	operation	
<u>Exan</u>	nple:	RETURN				

Examp	ole:	R

After Instruction: PC = TOS

	-	Detete L off	f thus work O		
RLC			f through C	arry	
Synta		RLCF f { $0 \le f \le 255$,o {,a}}		
Oper	ands:	0 ≤ 1 ≤ 255 d ∈ [0,1] a ∈ [0,1]			
Oper	ation:	$(f < n >) \rightarrow de$ $(f < 7 >) \rightarrow C,$ $(C) \rightarrow dest <$			
Statu	s Affected:	C, N, Z			
Enco	ding:	0011	01da fi	fff ffff	
Desc	escription: The contents of register 'f' are rotated one bit to the left through the Carry flag If 'd' is '0', the result is placed in W. If 'c is '1', the result is stored back in registe 'f' (default).				
If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).					
If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 23.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.					
		C	 regist 	ter f	
Word	s:	1			
Cycle	es:	1			
QC	ycle Activity:				
	Q1	Q2	Q3	Q4	
	Decode	Read register 'f'	Process Data	Write to destination	
l		Toglotor 1	Dulu	dootindion	
<u>Exan</u>	<u>iple:</u>	RLCF	REG, C	, 0	
	Before Instruc REG C After Instructio	= 1110 = 0	0110		
	REG W C	= 1110 = 1100 = 1			

RLNCF	Rotate Left f (No Carry)	RRCF	Rotate Right f through Carry
Syntax:	RLNCF f {,d {,a}}	Syntax:	RRCF f {,d {,a}}
Operands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$	Operands:	0 ≤ f ≤ 255 d ∈ [0,1] a ∈ [0,1]
Operation:	$(f < n >) \rightarrow dest < n + 1 >,$ $(f < 7 >) \rightarrow dest < 0 >$	Operation:	$(f < n >) \rightarrow dest < n - 1 >,$ $(f < 0 >) \rightarrow C,$ $(C) \rightarrow dest < 7 >$
Status Affected	N, Z	Status Affected:	
Encoding:	0100 01da ffff		-,,_
Description:	The contents of register 'f' are one bit to the left. If 'd' is '0', th is placed in W. If 'd' is '1', the r stored back in register 'f' (defa If 'a' is '0', the Access Bank is s	e result Description: esult is ult).	0011 00da ffff fff The contents of register 'f' are rotated one bit to the right through the Carry flag. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default)
	If 'a' is '1', the BSR is used to se GPR bank (default). If 'a' is '0' and the extended ins set is enabled, this instruction of	truction	register 'f' (default). If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).
	in Indexed Literal Offset Addre mode whenever f ≤ 95 (5Fh). S Section 23.2.3 "Byte-Oriente Bit-Oriented Instructions in I Literal Offset Mode" for detai	ssing See d and ndexed	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 23.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.
Words:	1		C register f
Cycles:	1		
Q Cycle Activit	v:	Words:	1
Q1		Q4 Cycles:	1
Decode	Read Process Wr	ite to Q Cycle Activit	•
	register 'f' Data dest	ination Q1	Q2 Q3 Q4
Example:	RLNCF REG, 1, 0	Decode	ReadProcessWrite toregister 'f'Datadestination
Before Ins	truction		
REG	= 1010 1011	Example:	RRCF REG, 0, 0
After Instru	uction	Before Inst	truction
REG	= 0101 0111	REG C	= 1110 0110 = 0
		After Instru	
		REG W C	= 1110 0110 = 0111 0011 = 0

RRNCF	Rotate Rig	ght f (No (Carry)			
Syntax:	RRNCF	f {,d {,a}}					
Operands:	0 ≤ f ≤ 255 d ∈ [0,1] a ∈ [0,1]						
Operation:		$(f < n >) \rightarrow dest < n - 1 >,$ $(f < 0 >) \rightarrow dest < 7 >$					
Status Affected:	N, Z						
Encoding:	0100	0100 00da ffff ffff					
Description:	one bit to t is placed i	The contents of register 'f' are rotated one bit to the right. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default).					
	If 'a' is '0', the Access Bank will be selected, overriding the BSR value. If 'a' is '1', then the bank will be selected as per the BSR value (default).						
	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 23.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.						
		► reg	gister	f	•		
Words:	1						
Cycles:	1						
Q Cycle Activity:							
Q1	Q2	Q3		(Q4		
Decode	Read register 'f'	Proces Data			te to nation		
Example 1:	RRNCF	REG, 1,	0				
Before Instruc REG		0111					
After Instructio REG		1011					
	= 1110	1011					
Example 2:	RRNCF		0				
Before Instruc	RRNCF		0				
	RRNCF tion = ? = 1101	REG, 0,	0				

	Set f					
x:	SETF f{,	a}				
nds:	0 ≤ f ≤ 255 a ∈ [0,1]					
tion:	$FFh\tof$					
Affected:	None					
ling:	0110	100a	ffff	ffff		
iption:			specified	l register		
	lf 'a' is '1', t	he BSR i				
If 'a' is 'o' and the extended instruction ope in Indexed Literal Offset Addressir mode whenever f ≤ 95 (5Fh). See Section 23.2.3 "Byte-Oriented an Bit-Oriented Instructions in Inde						
:	1					
3:	1	1				
cle Activity:						
Q1	Q2	Q3		Q4		
Decode	Read register 'f'			Write egister 'f'		
<u>ple:</u>	SETF	RE	G,1			
Before Instruct REG	= 5A	۱h				
	Inds: ition: Affected: ling: iption: iption: s: s: cle Activity: Q1 Decode ple: Before Instruct	unds: $0 \le f \le 255$ $a \in [0,1]$ ation:FFh \rightarrow fation:FFh \rightarrow fating:0110iption:The content are set to Fif 'a' is '0', tIf 'a' is '1', tGPR bankIf 'a' is '0' a set is enab in Indexed mode wherSection 23Bit-Oriente Literal Offss:1cle Activity:Q2DecodeRead register 'f'ple:SETFBefore Instruction	unds: $0 \le f \le 255$ $a \in [0,1]$ ation:FFh \rightarrow fa Affected:Noneling: 0110 $100a$ iption:The contents of the are set to FFh.If 'a' is '0', the Access If 'a' is '1', the BSR i GPR bank (default).If 'a' is '0' and the ex- set is enabled, this in in Indexed Literal Of mode whenever $f \le$ Section 23.2.3 "By Bit-Oriented Instru- Literal Offset Modes:1cle Activity:Q1Q2Q3DecodeRead register 'f'Proceple:SETFREGBefore InstructionREG	ands: $0 \le f \le 255$ $a \in [0,1]$ ation: FFh → f a Affected: None ling: 0110 100a ffff iption: The contents of the specified are set to FFh. If 'a' is '0', the Access Bank if f'a' is '1', the BSR is used to GPR bank (default). If 'a' is '0' and the extended is set is enabled, this instruction in Indexed Literal Offset Add mode whenever $f \le 95$ (5Fh) Section 23.2.3 "Byte-Orien Bit-Oriented Instructions in Literal Offset Mode" for defaults: s: 1 cle Activity: Q1 Q2 Q3 Decode Read Process r ple: SETF REG, 1 Before Instruction SETF REG, 1		

SLEEP	Enter Slee	ep Mode		SUBFWB	Subtract f from W with Borrow
Syntax:	SLEEP			Syntax:	SUBFWB f {,d {,a}}
Operands:	None			Operands:	$0 \le f \le 255$
Operation:	$00h \rightarrow WE$	DT,			d ∈ [0,1]
		postscaler,			a ∈ [0,1]
	$1 \rightarrow \underline{TO},$			Operation:	$(W) - (f) - (\overline{C}) \rightarrow dest$
o	$0 \rightarrow \overline{PD}$			Status Affected:	N, OV, C, DC, Z
Status Affected:	TO, PD			Encoding:	0101 01da ffff ffff
Encoding:	0000	0000 000		Description:	Subtract register 'f' and Carry flag
Description:		r-Down status	bit (PD) is tatus bit (TO)		(borrow) from W (2's complement
		Watchdog Tin			method). If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored in
		are cleared.			register 'f' (default).
	The proce	ssor is put into	Sleep mode		If 'a' is '0', the Access Bank is selected.
	with the os	scillator stoppe	d.		'a' is '1', the BSR is used to select the
Words:	1				GPR bank (default).
Cycles:	1				If 'a' is '0' and the extended instruction
Q Cycle Activity:					set is enabled, this instruction operates i Indexed Literal Offset Addressing mode
Q1	Q2	Q3	Q4		whenever f \leq 95 (5Fh). See
Decode	No	Process	Go to		Section 23.2.3 "Byte-Oriented and
	operation	Data	Sleep		Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.
				Words:	1
Example:	SLEEP			Cycles:	1
Before Instruc TO =	tion ?				
<u>10</u> = PD =	? ?			Q Cycle Activity: Q1	Q2 Q3 Q4
After Instruction	on			Decode	Read Process Write to
<u>TO</u> =	1†			Decode	register 'f' Data destination
PD =	0			Example 1:	SUBFWB REG, 1, 0
† If WDT causes	wake-up, this t	oit is cleared.		Before Instru	
				REG	= 3
				W C	= 2 = 1
				After Instruc	
				REG W	= FF = 2
				C Z	= 0
				Z N	= 0 = 1 ; result is negative
				Example 2:	SUBFWB REG, 0, 0
				Before Instru	
				REG W	= 2 = 5
				С	= 1
				After Instruc	
				REG W	= 2 = 3
				C Z	= 1 = 0
				N	= 0 ; result is positive
				Example 3:	SUBFWB REG, 1, 0
				Example 3:	SUBFWB REG, 1, 0

Before Instruction REG = W = C =

After Instruction

REG W C Z N = 1 = 2 = 0

= 0 = 2 = 1 = 1 = 0

; result is zero

SUBLW	Subtrac	Subtract W from Literal					
Syntax:	SUBLW	SUBLW k					
Operands:	$0 \le k \le 2$	255					
Operation:	k – (W)	\rightarrow W					
Status Affected:	N, OV, 0	C, DC, Z					
Encoding:	0000	1000 kkl	kk kkkk				
Description:		W is subtracted from the eight-bit literal 'k'. The result is placed in W.					
Words:	1						
Cycles:	1						
Q Cycle Activity:							
Q1	Q2	Q3	Q4				
Decode	Read literal 'k'	Process Data	Write to W				
Example 1:	SUBLW	02h					
Before Instruct W C After Instructio W C Z N	= 01h = ?						
Example 2:	SUBLW	02h					
Before Instruct W C After Instructio W C Z N	= 02h = ?	; result is zero					
Example 3:	SUBLW	02h					
Before Instruct W C After Instructio W C Z N	= 03h = ?	; (2's compleme ; result is negat					

SUBWF	Subtract W	from f			
Syntax:	SUBWF f	[,d {,a}}			
Operands:	$0 \leq f \leq 255$				
	d ∈ [0,1]				
Onerations	a ∈ [0,1]	1			
Operation:	$(f) - (W) \rightarrow 0$				
Status Affected:	N, OV, C, D(
Encoding:	0101	11da fff			
Description:	complement result is store	rom register 'f' method). If 'd' ed in W. If 'd' is ck in register 'f'	is '0', the '1', the result		
	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).				
	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \le 95$ (5Fh). See Section 23.2.3 "Byte-Oriented and				
		d Instructions et Mode" for d			
Words:	1		etalis.		
Cycles:	1				
2	I				
Q Cycle Activity: Q1	Q2	Q3	Q4		
Decode	Read	Process	Write to		
	register 'f'	Data	destination		
Example 1:	SUBWF	REG, 1, 0			
Before Instruc	tion				
REG W	= 3 = 2				
C	= ?				
After Instructio REG	on = 1				
W	= 2				
C Z	= 1 ; = 0	result is positiv	e		
Ň	= 0				
Example 2:	SUBWF	REG, 0, 0			
Before Instruc					
REG W	= 2 = 2				
С	= ?				
After Instructio REG	on = 2				
W	= 0				
C Z	= 1 ;1	result is zero			
Ň	= 0				
Example 3:	SUBWF	REG, 1, 0			
Before Instruc					
REG W	= 1 = 2				
С	= ?				
After Instructio		0'a aamalam	^+ \		
REG W	= FFh ;(= 2	2's complemer	n <i>)</i>		
C Z	= 0 ; = 0	result is negativ	/e		
Ň	= 0				

SUBWFB	Subtrac	t W from f v	vith Borr	ow		
Syntax:	SUBWF	3 f {,d {,a}]	}			
Operands:	$0 \le f \le 25$	55				
	d ∈ [0,1]					
	a ∈ [0,1]					
Operation:		$-(\overline{C}) \rightarrow des$	st			
Status Affected:	N, OV, C	, DC, Z		·		
Encoding:	0101	10da	ffff	ffff		
Description:		W and the (
		ister 'f' (2's c . If 'd' is '0', t				
		1' is '1', the r				
		er 'f' (default				
		', the Acces				
		', the BSR is nk (default).	s used to	select the		
	If 'a' is '0' and the extended instruction					
		abled, this in		•		
in Indexed Literal Offset Addressing mode whenever $f \le 95$ (5Fh). See						
	Section	23.2.3 "Byt	e-Orient	ed and		
Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.						
		offset Mode	" for deta	uls.		
Words:	1					
Cycles:	1					
Q Cycle Activity:	00	00		04		
Q1 Decode	Q2 Read	Q3 Proce	1	Q4 Write to		
Decode	register '			estination		
Example 1:	SUBWF	B REG, 1	, 0			
Before Instruct	tion	,				
REG	= 19h		L 1001)			
W C	= 0Dh = 1	(0000) 1101)			
After Instruction	n					
REG	= 0Ch) 1011)			
W C	= 0Dh = 1	(0000) 1101)			
Z N	= 0 = 0	· rocul	t is positi			
Example 2:	SUBWF			ve		
Before Instruc			, 0			
REG	= 1Bh	(000)	L 1011)			
W C	= 1Ah = 0	(0001	L 1010)			
After Instructio						
REG W						
	= 1Bh	(0001	L 1011)			
С						
C Z	= 1Bh = 00h = 1 = 1		t is zero			
C Z N	= 1Bh = 00h = 1 = 1 = 0	; resul	t is zero			
C Z N <u>Example 3:</u>	= 1Bh = 00h = 1 = 1 = 0 SUBWF	; resul	t is zero			
C Z N <u>Example 3:</u> Before Instruc REG	= 1Bh = 00h = 1 = 1 = 0 SUBWF tion = 03h	; resul B REG, 1 (0000	t is zero , 0) 0011)			
C Z N Example 3: Before Instruc REG W	= 1Bh = 00h = 1 = 0 SUBWF tion = 03h = 0Eh	; resul B REG, 1 (0000	t is zero			
C Z N <u>Example 3:</u> Before Instruc REG	= 1Bh = 00h = 1 = 0 SUBWF tion = 03h = 0Eh = 1	; resul B REG, 1 (0000	t is zero , 0) 0011)			
C Z N Example 3: Before Instruc REG W C	= 1Bh = 00h = 1 = 0 SUBWF tion = 03h = 0Eh = 1	; resul B REG, 1 (0000 (0000) (1111	t is zero , 0 0 0011) 0 1101) L 0100)			
C Z N Example 3: Before Instruct REG W C After Instructio	= 1Bh = 00h = 1 = 0 SUBWF tion = 03h = 0Eh = 1 n	; resul B REG, 1 (0000 (0000 (1111 ; [2's c	t is zero , 0 0 0011) 0 1101) L 0100) comp]			
C Z N Example 3: Before Instruct REG After Instruction REG	= 1Bh $= 00h$ $= 1$ $= 0$ SUBWF tion $= 0Sh$ $= 0Eh$ $= 1$ n $= F5h$ $= 0Eh$ $= 0$; resul B REG, 1 (0000 (0000 (1111 ; [2's c	t is zero , 0 0 0011) 0 1101) L 0100)			
C Z N Example 3: Before Instruct REG W C After Instruction REG	= 1Bh = 00h = 1 = 0 SUBWF tion = 03h = 0Eh = 1 n = F5h = 0Eh	; resul B REG, 1 (0000 (0000 (1111 ; [2's c (0000	t is zero , 0 0 0011) 0 1101) L 0100) comp]	ive		

SWAPF	Swap f					
Syntax:	SWAPF f	{,d {,a}}				
Operands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$					
Operation:	$(f<3:0>) \rightarrow$ $(f<7:4>) \rightarrow$					
Status Affected:	None					
Encoding:	0011	10da	fff	f f	fff	
Description:	'f' are excha is placed in	The upper and lower nibbles of register f' are exchanged. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed in register 'f' (default).				
	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).					
	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 23.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.					
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3	1	Q	4	
Decode	Read register 'f'	Proce Data		Write destina		
<u>Example:</u> Before Instruc REG After Instructio REG	tion = 53h	2EG, 1,	0			

Table Read (Continued)

*+ ;

TBLRD	Table Read						
Syntax:	TBLRD (*;	*+; *-; +	-*)				
Operands:	None	None					
Operation:	if TBLRD *, (Prog Mem (TBLPTR)) \rightarrow TABLAT; TBLPTR – No Change if TBLRD *+, (Prog Mem (TBLPTR)) \rightarrow TABLAT; (TBLPTR) + 1 \rightarrow TBLPTR if TBLRD *-, (Prog Mem (TBLPTR)) \rightarrow TABLAT; (TBLPTR) – 1 \rightarrow TBLPTR if TBLRD +*, (TBLPTR) + 1 \rightarrow TBLPTR; (Prog Mem (TBLPTR)) \rightarrow TABLAT						
Status Affected:	None						
Encoding:	0000	0000	000	00	10nn nn=0 * =1 *+ =2 *- =3 +*		
Description:	This instruction is used to read the contents of Program Memory (P.M.). To address the program memory, a pointer called Table Pointer (TBLPTR) is used.				dress the I Table		
	The TBLPT each byte in has a 2-Mby	the pro	ogram me	emory			
	TBLPTR[(-	Program [®] I	Memo	ory Word		
	TBLPTR[(Most Sign Program I				
	The TBLRD of TBLPTR			nodify	the value		
	no chang	е					
	 post-incre 	ement					
	 post-decr 						
	 pre-increi 	nent					
Words:	1						
Cycles:	2						
Q Cycle Activity	:						
Q1	Q2		Q3	1	Q4		
Decode	No operation	op	No peration	ор	No eration		

Before Instruction		
TABLAT	=	55h
TBLPTR	=	00A356h
MEMORY(00A356h)	=	34h
After Instruction		
TABLAT	=	34h
TRI PTR	_	00A357h

TBLRD

TBLRD

Example 1:

TBLPTR			=	00A357h
Example 2:	TBLRD	+*	;	
Before Instruction	on			
TABLAT TBLPTR MEMORY(MEMORY(01A357h) 01A358h)		= = =	AAh 01A357h 12h 34h
After Instruction TABLAT TBLPTR			= =	34h 01A358h

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No operation (Read Program

Memory)

No

operation

No operation (Write TABLAT)

No

operation

TBLWT	Table Wri	te					
Syntax:	TBLWT ('	*; *+; *-; +*	r)				
Operands:	None						
Operation:	if TBLWT*, (TABLAT) \rightarrow Holding Register; TBLPTR – No Change if TBLWT*+, (TABLAT) \rightarrow Holding Register; (TBLPTR) + 1 \rightarrow TBLPTR if TBLWT*-, (TABLAT) \rightarrow Holding Register; (TBLPTR) – 1 \rightarrow TBLPTR if TBLWT+*, (TBLPTR) + 1 \rightarrow TBLPTR;						
	(TABLAT)						
Status Affected:	None						
Encoding:	0000	0000	0000	11nn			
-				nn=0 *			
				=1 *+ =2 *-			
				=2 ^-			
	8 holding registers the TABLAT is without to. The holding registers are used to program the contents of Program Meters (P.M.). (Refer to Section 5.0 "Memory Organization" for additional details programming Flash memory.) The TBLPTR (a 21-bit pointer) point each byte in the program memory. TBLPTR has a 2-Mbyte address ran The LSb of the TBLPTR selects while byte of the program memory location access. TBLPTR[0] = 0: Least Significant of Program Memory Memory address ran the LPTR[0] = 1: Most Significant of Program Memory Me						
	The TBLW value of T • no char • post-inc	BLPTR as		odify the			
	 post-de 	crement					
	 pre-incr 	rement					
Words:	1						
Cycles:	2						
Q Cycle Activity:	C :			<i></i>			
	Q1	Q2	Q3	Q4			
	Decode	No operation	No operation	No operation			
	No operation	No operation (Read TABLAT)	No	No			

TBLWT	Table Write (C	ontin	ued)
Example 1:	TBLWT *+;		
Before Instru TABLAT		=	55h
TBLPTF HOLDIN (00A356	IG REGISTER	=	00A356h FFh
•	ions (table write o	= comp	
TABLAT		=	55h
TBLPTF HOLDIN	REGISTER	=	00A357h
(00A356	Sh)	=	55h
Example 2:	TBLWT +*;		
Before Instru	ction		
		= =	34h 01389Ah
(01389A		=	FFh
(01389E	Bh)	=	FFh
After Instruct	ion (table write co	omple	etion)
TABLAT		=	34h
	{ IG REGISTER	=	01389Bh
(01389A		=	FFh
(01389E		=	34h

Register)

TSTFSZ	Test f, Skip	o if O			
Syntax:	TSTFSZ f {	,a}			
Operands:	$\begin{array}{l} 0 \leq f \leq 255 \\ a \in \ [0,1] \end{array}$				
Operation:	skip if f = 0				
Status Affected:	None				
Encoding:	0110	011a fff	f ffff		
Description: If 'f' = 0, the next instruction fetched during the current instruction executi is discarded and a NOP is executed, making this a two-cycle instruction.					
	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).				
	set is enabl in Indexed mode wher Section 23 Bit-Oriente	nd the extende ed, this instruct Literal Offset A ever f ≤ 95 (5F .2.3 "Byte-Ori d Instructions set Mode" for	tion operates ddressing h). See ented and s in Indexed		
Words:	1				
Cycles: Q Cycle Activity:		rcles if skip and a 2-word instru			
Q Oycle Activity. Q1	Q2	Q3	Q4		
Decode	Read	Process	No		
	register 'f'	Data	operation		
If skip:		0.0	<u>.</u>		
Q1 No	Q2 No	Q3 No	Q4 No		
operation	operation	operation	operation		
If skip and followe	. ·				
Q1	Q2	Q3	Q4		
No	No	No	No		
operation	operation	operation	operation		
No	No	No	No		
operation	operation	operation	operation		
operation	HERE T	ISTFSZ CNT			
Example: Before Instruc	HERE T NZERO ZERO	ISTFSZ CNT :	, 1		
Example: Before Instruc PC	HERE T NZERO ZERO Ction = Ad	ISTFSZ CNT	, 1		
Example: Before Instruc	HERE NZERO ZERO ction = Ad on = 00	TSTFSZ CNT : : dress (here)	, 1		

XOR	LW	Exclusive	Exclusive OR Literal with W				
Synt	ax:	XORLW	k				
Oper	rands:	$0 \le k \le 25$	5				
Oper	ration:	(W) .XOR	l. k \rightarrow W				
Statu	is Affected:	N, Z					
Enco	oding:	0000	1010	kkkk	kkkk		
Desc	cription:	The contents of W are XORed with the 8-bit literal 'k'. The result is placed in W.					
Word	ds:	1	1				
Cycle	es:	1					
QC	ycle Activity:						
	Q1	Q2	Q3		Q4		
	Decode	Read literal 'k'	Proce: Data		Write to W		
<u>Exar</u>	Example: XORLW 0AFh						
	Before Instruction W = B5h After Instruction						

W = 1Ah

XORWF	Exclusive OR W with f					
Syntax:	XORWF	f {,d {,a}}				
Operands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$					
Operation:	(W) .XOR.	(W) .XOR. (f) \rightarrow dest				
Status Affected:	N, Z					
Encoding:	0001	10da ff:	ff ffff			
Description:	register 'f'. in W. If 'd' is	DR the content If 'd' is 'o', the r s '1', the result ter 'f' (default).	esult is stored is stored back			
	,	he Access Bai he BSR is use (default).				
	set is enabl in Indexed mode wher Section 23 Bit-Oriente	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 23.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.				
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3	Q4			
Decode	Read register 'f'	Process Data	Write to destination			
Example:		REG, 1, 0				
Before Instruc REG W	tion = AFh = B5h					
After Instructio REG W	on = 1Ah = B5h					

23.2 Extended Instruction Set

In addition to the standard 75 instructions of the PIC18 instruction set, the PIC18F85J90 family family of devices also provide an optional extension to the core CPU functionality. The added features include eight additional instructions that augment Indirect and Indexed Addressing operations and the implementation of Indexed Literal Offset Addressing for many of the standard PIC18 instructions.

The additional features of the extended instruction set are enabled by default on unprogrammed devices. Users must properly set or clear the XINST Configuration bit during programming to enable or disable these features.

The instructions in the extended set can all be classified as literal operations, which either manipulate the File Select Registers, or use them for Indexed Addressing. Two of the instructions, ADDFSR and SUBFSR, each have an additional special instantiation for using FSR2. These versions (ADDULNK and SUBULNK) allow for automatic return after execution.

The extended instructions are specifically implemented to optimize re-entrant program code (that is, code that is recursive or that uses a software stack) written in high-level languages, particularly C. Among other things, they allow users working in high-level languages to perform certain operations on data structures more efficiently. These include:

- Dynamic allocation and deallocation of software stack space when entering and leaving subroutines
- Function Pointer invocation
- Software Stack Pointer manipulation
- Manipulation of variables located in a software stack

A summary of the instructions in the extended instruction set is provided in Table 23-3. Detailed descriptions are provided in **Section 23.2.2 "Extended Instruction Set**". The opcode field descriptions in Table 23-1 (page 296) apply to both the standard and extended PIC18 instruction sets.

Note: The instruction set extension and the Indexed Literal Offset Addressing mode were designed for optimizing applications written in C; the user may likely never use these instructions directly in assembler. The syntax for these commands is provided as a reference for users who may be reviewing code that has been generated by a compiler.

23.2.1 EXTENDED INSTRUCTION SYNTAX

Most of the extended instructions use indexed arguments, using one of the File Select Registers and some offset to specify a source or destination register. When an argument for an instruction serves as part of Indexed Addressing, it is enclosed in square brackets ("[]"). This is done to indicate that the argument is used as an index or offset. The MPASM[™] Assembler will flag an error if it determines that an index or offset value is not bracketed.

When the extended instruction set is enabled, brackets are also used to indicate index arguments in byte-oriented and bit-oriented instructions. This is in addition to other changes in their syntax. For more details, see Section 23.2.3.1 "Extended Instruction Syntax with Standard PIC18 Commands".

Note: In the past, square brackets have been used to denote optional arguments in the PIC18 and earlier instruction sets. In this text and going forward, optional arguments are denoted by braces ("{ }").

Mnemonic, Operands		Description	Cycles			Word Status		
		Description	Cycles	MSb			LSb	Affected
ADDFSR	f, k	Add literal to FSR	1	1110	1000	ffkk	kkkk	None
ADDULNK	k	Add literal to FSR2 and return	2	1110	1000	11kk	kkkk	None
CALLW		Call subroutine using WREG	2	0000	0000	0001	0100	None
MOVSF	z _s , f _d	Move z _s (source) to 1st word	2	1110	1011	0zzz	ZZZZ	None
		f _d (destination) 2nd word		1111	ffff	ffff	ffff	
MOVSS	z _s , z _d	Move z _s (source) to 1st word	2	1110	1011	lzzz	ZZZZ	None
		z _d (destination) 2nd word		1111	xxxx	XZZZ	ZZZZ	
PUSHL	k	Store literal at FSR2,	1	1110	1010	kkkk	kkkk	None
		decrement FSR2						
SUBFSR	f, k	Subtract literal from FSR	1	1110	1001	ffkk	kkkk	None
SUBULNK	k	Subtract literal from FSR2 and	2	1110	1001	11kk	kkkk	None
		return						

TABLE 23-3: EXTENSIONS TO THE PIC18 INSTRUCTION SET

23.2.2 EXTENDED INSTRUCTION SET

ADD	FSR	Add Literal to FSR				
Synta	ax:	ADDFSR	f, k			
Oper	ands:	$0 \le k \le 63$	$0 \le k \le 63$			
		f ∈ [0, 1,	f ∈ [0, 1, 2]			
Oper	ation:	$FSR(f) + k \rightarrow FSR(f)$				
Status Affected: None						
Enco	ding:	1110	1000	ffkk	kkkk	
Desc	ription:	The 6-bit	The 6-bit literal 'k' is added to the			
		contents of	of the FSF	R specif	ied by 'f'.	
Word	ls:	1	1			
Cycle	es:	1				
QC	ycle Activity:					
	Q1	Q2	Q3		Q4	
	Decode	Read	Proces	ss	Write to	
		literal 'k'	Data	L	FSR	

Example: ADD

ADDFSR 2, 23h

Before Instru						
FSR2	=	03FFh				
After Instruction						
FSR2	=	0422h				

ADD	ULNK	Add Literal to FSR2 and Return					
Synta	ax:	ADDULN	(k				
Oper	ands:	$0 \le k \le 63$					
Oper	ation:	FSR2 + k	\rightarrow FSR2,				
		$(TOS) \rightarrow I$	PC				
Statu	s Affected:	None					
Enco	ding:	1110	1110 1000 11kk kkkk				
Desc	ription:	The 6-bit literal 'k' is added to the contents of FSR2. A RETURN is then executed by loading the PC with the TOS.					
		The instru execute; a the secon	NOP is pe		cycles to ned during		
		case of the	e ADDFSR 3 (binary '	instr	s a special uction, it operates		
Word	ls:	1					
Cycle	es:	2					
QC	ycle Activity:						
	Q1	Q2	Q3		Q4		
	Decode	Read literal 'k'	Proces Data	S	Write to FSR		
	No	No	No		No		
	Operation	Operation	Operatio	on	Operation		
Example: ADDULNK 23h							

Example:	AD	DULNK 2			
Before Instruction					
FSR2	=	03FFh			
PC	=	0100h			
After Instruc	tion				
FSR2	=	0422h			
PC	=	(TOS)			

Note: All PIC18 instructions may take an optional label argument preceding the instruction mnemonic for use in symbolic addressing. If a label is used, the instruction format then becomes: {label} instruction argument(s).

CAL	LW	Subroutine	e Call Using	WRE	G		
Synt	ax:	CALLW					
Oper	rands:	None					
Oper	ration:	$\begin{array}{l} (W) \rightarrow PCL \\ (PCLATH) \end{array}$	$(PC + 2) \rightarrow TOS,$ $(W) \rightarrow PCL,$ $(PCLATH) \rightarrow PCH,$ $(PCLATU) \rightarrow PCU$				
Statu	us Affected:	None					
Enco	oding:	0000	0000 0	001	0100		
Desc	pription	First, the return address (PC + 2) is pushed onto the return stack. Next, the contents of W are written to PCL; the existing value is discarded. Then, the contents of PCLATH and PCLATU are latched into PCH and PCU, respectively. The second cycle is executed as a NOP instruction while the new next instruction is fetched. Unlike CALL, there is no option to					
			L, there is no STATUS or E	•	n to		
Word	ds:	1					
Cycle	es:	2					
QC	ycle Activity:						
	Q1	Q2	Q3		Q4		
	Decode	Read WREG	Push PC to stack		No peration		
	No	No	No		No		
	operation	operation	operation	op	eration		
<u>Exar</u>	nple: Before Instruc PCLATH PCLATH W After Instructio PC	= address = 10h = 00h = 06h	CALLW (HERE) h	2)			

	~-					
MOV		Move Inde				
Synta	ax:		z _s], f _d			
Oper	ands:	$0 \le z_s \le 12^{\circ}$ $0 \le f_d \le 408^{\circ}$				
Oper	ation:	((FSR2) + 2	$z_s) \rightarrow f_d$			
Statu	s Affected:	None				
	ding: ord (source) vord (destin.)	1110 1111	1011 ffff	0zz fff	ъ	
Desc	ription:	The conten	ts of the s	ource	e register are	
		moved to destination register 'f _d '. The actual address of the source register is determined by adding the 7-bit literal offset 'z _s ', in the first word, to the value of FSR2. The address of the destination register is specified by the 12-bit literal 'f _d ' in the second word. Both addresses can be anywhere in the 4096-byte data space (000h to FFFh).				
		The MOVSF PCL, TOSU destination	J, TOSH o		not use the SL as the	
		If the result an Indirect value returi	Addressin	g reg	-	
Word	s:	2				
Cycle	es:	2				
Q C	vcle Activity:					
	Q1	Q2	Q3		Q4	
	Decode	Determine	Determi	-	Read	
	Decede	source addr	source a	ddr	source reg	
	Decode	No operation	No operatio	on	Write register 'f'	
		No dummy			(dest)	
		read				
Exam	<u>iple:</u>	MOVSF	[05h], F	REG2		
	Before Instruc	tion				
FSR2 = 80h Contents of 85h = 33h REG2 = 11h						
	After Instruction	on				
	FSR2 Contents	= 80	h			
	of 85h REG2	= 33 = 33				

MOVSS	Move Inde	Move Indexed to Indexed					
Syntax:	MOVSS	[z _s], [z _d]					
Operands:	0	$0 \le z_s \le 127$ $0 \le z_d \le 127$					
Operation:	((FSR2) +	$((FSR2) + z_s) \rightarrow ((FSR2) + z_d)$					
Status Affected:	None						
Encoding: 1st word (source) 2nd word (dest.) Description	11101011 $1zzz$ $zzzz_s$ 1111 $xxxx$ $xzzz$ $zzzz_d$ The contents of the source register are moved to the destination register. The addresses of the source and destination registers are determined by adding the 7-bit literal offsets, ' z_s ' or ' z_d ', respectively, to the value of FSR2. Both registers can be located anywhere in the 4096-byte data memory space						
	The MOVSS instruction cannot use the PCL, TOSU, TOSH or TOSL as the destination register.						
	If the resultant source address points to an Indirect Addressing register, the value returned will be 00h. If the resultant destination address points to an Indirect Addressing register, the instruction will execute as a NOP.						
Words:	2						
Cycles:	2						
Q Cvcle Activity:							

Cycles.	2
Q Cycle Activity:	

_	Q1	Q2	Q3	Q4
	Decode	Determine	Determine	Read
		source addr	source addr	source reg
	Decode	Determine dest addr	Determine dest addr	Write to dest reg

Example:	<u>movss</u>		[06h]
Before Instructio	on =	80h	
Contents of 85h Contents	=	33h	
of 86h After Instruction	=	11h	
FSR2 Contents	=	80h	
of 85h Contents	=	33h	
of 86h	=	33h	

PUSHL	Store Literal	at FSR	2, Decr	ement F	SR2	
Syntax:	PUSHL k					
Operands:	$0 \le k \le 255$	$0 \le k \le 255$				
Operation:	$k \rightarrow (FSR2),$ FSR2 – 1 \rightarrow FSR2					
Status Affected:	None					
Encoding:	1111	1010	kkk]	k kk	kk	
	memory address specified by FSR2. FSR2 is decremented by 1 after the operation. This instruction allows users to push					
Words:	values onto	a sonwa	re staci	к.		
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	C	03	Q4		
Decode	Read 'k'	Proc da		Write destina		
Example:	PUSHL 08		ta	destina		
Before Instrue	ction					

Before Instruction FSR2H:FSR2L Memory (01ECh)	= =	01ECh 00h
After Instruction FSR2H:FSR2L Memory (01ECh)	= =	01EBh 08h

SUBULNK k

 $\mathsf{FSR2}-\mathsf{k}\to\mathsf{FSR2},$

 $0 \leq k \leq 63$

Subtract Literal from FSR2 and Return

SUB	UBFSR Subtract Literal from FSR						
Synta	ax:	SUBFSR	f, k				
Oper	ands:	$0 \le k \le 63$	$0 \le k \le 63$				
		f∈[0,1,	2]				
Oper	ation:	FSRf – k	\rightarrow FSRf				
Statu	Status Affected: None						
Enco	ding:	1110	1001	ffkk	kkkk		
Desc	ription:	The 6-bit I the conter by 'f'.					
Word	ls:	1					
Cycle	es:	1	1				
QC	ycle Activity:						
	Q1	Q2	Q3	i	Q4		
	Decode	Read	Proce	SS	Write to		
		register 'f'	Data	a d	estination		
<u>Exan</u>		SUBFSR :	2, 23h				
	Before Instruction						

			$(TOS) \rightarrow PC$					
	Statu	s Affected:	None					
	Enco	ding:	1110	100)1	11kk		kkkk
_	Desc	ription:	The 6-bit literal 'k' is subtracted from th contents of the FSR2. A RETURN is the executed by loading the PC with the TOS.					is then
			The instru- execute; a second cy	NOP				
			This may be thought of as a special case of the SUBFSR instruction, where f = 3 (binary '11'); it operates only on FSR2.					
	Word	s:	1					
	Cycle	es:	2					
	QC	cle Activity:						
		Q1	Q2		C	23		Q4
		Decode	Read register	'f'	Proc Da			/rite to stination
		No	No		Ν	0		No
		Operation	Operatio	n	Oper	ation	Op	peration

Example: SUBULNK 23h

SUBULNK

Operands:

Operation:

Syntax:

Before Instruction							
FSR2	=	03FFh					
PC	=	0100h					
After Instruct	ion						
FSR2	=	03DCh					
PC	=	(TOS)					

Before Instruction 03FFh FSR2 = After Instruction FSR2 03DCh =

23.2.3 BYTE-ORIENTED AND BIT-ORIENTED INSTRUCTIONS IN INDEXED LITERAL OFFSET MODE

Note:	Enabling the PIC18 instruction set exten-						
	sion may cause legacy applications to						
	behave erratically or fail entirely.						

In addition to eight new commands in the extended set, enabling the extended instruction set also enables Indexed Literal Offset Addressing (Section 5.6.1 "Indexed Addressing with Literal Offset"). This has a significant impact on the way that many commands of the standard PIC18 instruction set are interpreted.

When the extended set is disabled, addresses embedded in opcodes are treated as literal memory locations: either as a location in the Access Bank (a = 0) or in a GPR bank designated by the BSR (a = 1). When the extended instruction set is enabled and a = 0, however, a file register argument of 5Fh or less is interpreted as an offset from the pointer value in FSR2 and not as a literal address. For practical purposes, this means that all instructions that use the Access RAM bit as an argument – that is, all byte-oriented and bit-oriented instructions, or almost half of the core PIC18 instructions – may behave differently when the extended instruction set is enabled.

When the content of FSR2 is 00h, the boundaries of the Access RAM are essentially remapped to their original values. This may be useful in creating backward-compatible code. If this technique is used, it may be necessary to save the value of FSR2 and restore it when moving back and forth between C and assembly routines in order to preserve the Stack Pointer. Users must also keep in mind the syntax requirements of the extended instruction set (see Section 23.2.3.1 "Extended Instruction Syntax with Standard PIC18 Commands").

Although the Indexed Literal Offset mode can be very useful for dynamic stack and pointer manipulation, it can also be very annoying if a simple arithmetic operation is carried out on the wrong register. Users who are accustomed to the PIC18 programming must keep in mind that, when the extended instruction set is enabled, register addresses of 5Fh or less are used for Indexed Literal Offset Addressing.

Representative examples of typical byte-oriented and bit-oriented instructions in the Indexed Literal Offset mode are provided on the following page to show how execution is affected. The operand conditions shown in the examples are applicable to all instructions of these types.

23.2.3.1 Extended Instruction Syntax with Standard PIC18 Commands

When the extended instruction set is enabled, the file register argument 'f' in the standard byte-oriented and bit-oriented commands is replaced with the literal offset value 'k'. As already noted, this occurs only when 'f' is less than or equal to 5Fh. When an offset value is used, it must be indicated by square brackets ("[]"). As with the extended instructions, the use of brackets indicates to the compiler that the value is to be interpreted as an index or an offset. Omitting the brackets, or using a value greater than 5Fh within the brackets, will generate an error in the MPASM Assembler.

If the index argument is properly bracketed for Indexed Literal Offset Addressing, the Access RAM argument is never specified; it will automatically be assumed to be '0'. This is in contrast to standard operation (extended instruction set disabled), when 'a' is set on the basis of the target address. Declaring the Access RAM bit in this mode will also generate an error in the MPASM Assembler.

The destination argument 'd' functions as before.

In the latest versions of the MPASM Assembler, language support for the extended instruction set must be explicitly invoked. This is done with either the command line option, $/_{Y}$, or the PE directive in the source listing.

23.2.4 CONSIDERATIONS WHEN ENABLING THE EXTENDED INSTRUCTION SET

It is important to note that the extensions to the instruction set may not be beneficial to all users. In particular, users who are not writing code that uses a software stack may not benefit from using the extensions to the instruction set.

Additionally, the Indexed Literal Offset Addressing mode may create issues with legacy applications written to the PIC18 assembler. This is because instructions in the legacy code may attempt to address registers in the Access Bank below 5Fh. Since these addresses are interpreted as literal offsets to FSR2 when the instruction set extension is enabled, the application may read or write to the wrong data addresses.

When porting an application to the PIC18F85J90 family family, it is very important to consider the type of code. A large, re-entrant application that is written in C and would benefit from efficient compilation will do well when using the instruction set extensions. Legacy applications that heavily use the Access Bank will most likely not benefit from using the extended instruction set.

ADD	WF	ADD W to I (Indexed L		iset n	node	e)
Synta	ax:	ADDWF	[k] {,d}			
Oper	ands:	$\begin{array}{l} 0 \leq k \leq 95 \\ d \in \ [0,1] \end{array}$				
Oper	ration:	(W) + ((FSF	R2) + k) -	→ des	t	
Statu	is Affected:	N, OV, C, D	C, Z			
Enco	oding:	0010	01d0	kkk	k	kkkk
Desc	cription:	The content contents of FSR2, offse	the regis	ter ind	dicat	
		If 'd' is '0', the result is stored in W. If 'd is '1', the result is stored back in register 'f' (default).				
Word	ds:	1				
Cycle	es:	1				
QC	ycle Activity:					
	Q1	Q2	Q3		0	Q4
	Decode	Read 'k'	Proce: Data			/rite to stination
<u>Exar</u>	nple:	ADDWF	[OFST] ,	0		
	Before Instruction W OFST FSR2 Contents of 0A2Ch	on = = = =	17h 2Ch 0A00h 20h			
	After Instruction W Contents of 0A2Ch		37h 20h			
		_	2011			

BSF	Bit Set Inde (Indexed L	exed iteral Offset i	node)						
Syntax:	BSF [k], b								
Operands:	$\begin{array}{l} 0 \leq f \leq 95 \\ 0 \leq b \leq 7 \end{array}$								
Operation:	$1 \rightarrow ((FSR2) + k) < b >$								
Status Affected:	None								
Encoding:	1000	bbb0 kki	kk kkkk						
Description:		register indic e value 'k', is s	ated by FSR2, set.						
Words:	1								
Cycles:	1								
Q Cycle Activity:									
Q1	Q2	Q3	Q4						
Decode	Read register 'f'	Process Data	Write to destination						
Example:	BSF [FLAG_OFST]	, 7						
Before Instructi FLAG_OF FSR2 Contents of 0A0Ah After Instruction	ST = = =	0Ah 0A00h 55h							
Contents of 0A0Ah	=	D5h							
SETF	Set Indexe (Indexed L	d iteral Offset ı	node)						
SETF Syntax:			node)						
	(Indexed L		mode)						
Syntax:	(Indexed L SETF [k]	iteral Offset ı	node)						
Syntax: Operands:	(Indexed L SETF [k] $0 \le k \le 95$	iteral Offset ı	mode)						
Syntax: Operands: Operation:	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS	iteral Offset ı							
Syntax: Operands: Operation: Status Affected:	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The content	iteral Offset (6R2) + k) 1000 kk	kk kkkk er indicated by						
Syntax: Operands: Operation: Status Affected: Encoding:	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The content	BR2) + k)	kk kkkk er indicated by						
Syntax: Operands: Operation: Status Affected: Encoding: Description:	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The content FSR2, offset	BR2) + k)	kk kkkk er indicated by						
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words:	(Indexed L SETF $[k]$ $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The content FSR2, offset 1	BR2) + k)	kk kkkk er indicated by						
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles:	(Indexed L SETF $[k]$ $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The content FSR2, offset 1	BR2) + k)	kk kkkk er indicated by						
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity:	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The content FSR2, offset 1 1	BR2) + k) 1000 kk ts of the regist at by 'k', are se	kk kkkk er indicated by et to FFh.						
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The content FSR2, offset 1 1 Q2 Read 'k'	BR2) + k) 1000 kk ts of the regist at by 'k', are so Q3 Process Data	kk kkkk er indicated by et to FFh. Q4 Write						
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode Example:	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The content FSR2, offset 1 1 2 Read 'k'	BR2) + k) 1000 kk ts of the regist et by 'k', are se Q3 Process	kk kkkk er indicated by et to FFh. Q4 Write						
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode Example: Before Instructi	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The content FSR2, offset 1 1 Q2 Read 'k' SETF [on = 2C	BR2) + k) 1000 kk sof the regist of the regist by 'k', are so Q3 Process Data COFST] h	kk kkkk er indicated by et to FFh. Q4 Write						
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode Example: Before Instruction	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The content FSR2, offset 1 1 Q2 Read 'k' SETF [on = 2C	iteral Offset i SR2) + k) 1000 kk ts of the regist et by 'k', are se Q3 Process Data "OFST]	kk kkkk er indicated by et to FFh. Q4 Write						
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode Example: Before Instruction OFST FSR2	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The content FSR2, offset 1 1 Q2 Read 'k' SETF [on = 2C = 0A = 00	iteral Offset i SR2) + k) 1000 kk ts of the regist ts of the regist to by 'k', are se Q3 Process Data COFST] h 00h	kk kkkk er indicated by et to FFh. Q4 Write						

23.2.5 SPECIAL CONSIDERATIONS WITH MICROCHIP MPLAB® IDE TOOLS

The latest versions of Microchip's software tools have been designed to fully support the extended instruction set for the PIC18F85J90 family family. This includes the MPLAB C18 C Compiler, MPASM assembly language and MPLAB Integrated Development Environment (IDE).

When selecting a target device for software development, MPLAB IDE will automatically set default Configuration bits for that device. The default setting for the XINST Configuration bit is '0', disabling the extended instruction set and Indexed Literal Offset Addressing. For proper execution of applications developed to take advantage of the extended instruction set, XINST must be set during programming.

To develop software for the extended instruction set, the user must enable support for the instructions and the Indexed Addressing mode in their language tool(s). Depending on the environment being used, this may be done in several ways:

- A menu option or dialog box within the environment that allows the user to configure the language tool and its settings for the project
- A command line option
- A directive in the source code

These options vary between different compilers, assemblers and development environments. Users are encouraged to review the documentation accompanying their development systems for the appropriate information.

24.0 DEVELOPMENT SUPPORT

The PIC[®] microcontrollers are supported with a full range of hardware and software development tools:

- Integrated Development Environment
 - MPLAB[®] IDE Software
- Assemblers/Compilers/Linkers
 - MPASM[™] Assembler
 - MPLAB C18 and MPLAB C30 C Compilers
 - MPLINK[™] Object Linker/
 - MPLIB™ Object Librarian
 - MPLAB ASM30 Assembler/Linker/Library
- Simulators
 - MPLAB SIM Software Simulator
- Emulators
 - MPLAB ICE 2000 In-Circuit Emulator
 - MPLAB REAL ICE™ In-Circuit Emulator
- In-Circuit Debugger
 - MPLAB ICD 2
- Device Programmers
 - PICSTART[®] Plus Development Programmer
 - MPLAB PM3 Device Programmer
 - PICkit[™] 2 Development Programmer
- Low-Cost Demonstration and Development Boards and Evaluation Kits

24.1 MPLAB Integrated Development Environment Software

The MPLAB IDE software brings an ease of software development previously unseen in the 8/16-bit microcontroller market. The MPLAB IDE is a Windows[®] operating system-based application that contains:

- A single graphical interface to all debugging tools
 - Simulator
 - Programmer (sold separately)
 - Emulator (sold separately)
 - In-Circuit Debugger (sold separately)
- · A full-featured editor with color-coded context
- A multiple project manager
- Customizable data windows with direct edit of contents
- High-level source code debugging
- Visual device initializer for easy register initialization
- Mouse over variable inspection
- Drag and drop variables from source to watch windows
- Extensive on-line help
- Integration of select third party tools, such as HI-TECH Software C Compilers and IAR C Compilers

The MPLAB IDE allows you to:

- Edit your source files (either assembly or C)
- One touch assemble (or compile) and download to PIC MCU emulator and simulator tools (automatically updates all project information)
- Debug using:
 - Source files (assembly or C)
 - Mixed assembly and C
 - Machine code

MPLAB IDE supports multiple debugging tools in a single development paradigm, from the cost-effective simulators, through low-cost in-circuit debuggers, to full-featured emulators. This eliminates the learning curve when upgrading to tools with increased flexibility and power.

24.2 MPASM Assembler

The MPASM Assembler is a full-featured, universal macro assembler for all PIC MCUs.

The MPASM Assembler generates relocatable object files for the MPLINK Object Linker, Intel[®] standard HEX files, MAP files to detail memory usage and symbol reference, absolute LST files that contain source lines and generated machine code and COFF files for debugging.

The MPASM Assembler features include:

- Integration into MPLAB IDE projects
- User-defined macros to streamline assembly code
- Conditional assembly for multi-purpose source files
- Directives that allow complete control over the assembly process

24.3 MPLAB C18 and MPLAB C30 C Compilers

The MPLAB C18 and MPLAB C30 Code Development Systems are complete ANSI C compilers for Microchip's PIC18 and PIC24 families of microcontrollers and the dsPIC30 and dsPIC33 family of digital signal controllers. These compilers provide powerful integration capabilities, superior code optimization and ease of use not found with other compilers.

For easy source level debugging, the compilers provide symbol information that is optimized to the MPLAB IDE debugger.

24.4 MPLINK Object Linker/ MPLIB Object Librarian

The MPLINK Object Linker combines relocatable objects created by the MPASM Assembler and the MPLAB C18 C Compiler. It can link relocatable objects from precompiled libraries, using directives from a linker script.

The MPLIB Object Librarian manages the creation and modification of library files of precompiled code. When a routine from a library is called from a source file, only the modules that contain that routine will be linked in with the application. This allows large libraries to be used efficiently in many different applications.

The object linker/library features include:

- Efficient linking of single libraries instead of many smaller files
- Enhanced code maintainability by grouping related modules together
- Flexible creation of libraries with easy module listing, replacement, deletion and extraction

24.5 MPLAB ASM30 Assembler, Linker and Librarian

MPLAB ASM30 Assembler produces relocatable machine code from symbolic assembly language for dsPIC30F devices. MPLAB C30 C Compiler uses the assembler to produce its object file. The assembler generates relocatable object files that can then be archived or linked with other relocatable object files and archives to create an executable file. Notable features of the assembler include:

- Support for the entire dsPIC30F instruction set
- · Support for fixed-point and floating-point data
- · Command line interface
- Rich directive set
- Flexible macro language
- MPLAB IDE compatibility

24.6 MPLAB SIM Software Simulator

The MPLAB SIM Software Simulator allows code development in a PC-hosted environment by simulating the PIC MCUs and dsPIC[®] DSCs on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a comprehensive stimulus controller. Registers can be logged to files for further run-time analysis. The trace buffer and logic analyzer display extend the power of the simulator to record and track program execution, actions on I/O, most peripherals and internal registers.

The MPLAB SIM Software Simulator fully supports symbolic debugging using the MPLAB C18 and MPLAB C30 C Compilers, and the MPASM and MPLAB ASM30 Assemblers. The software simulator offers the flexibility to develop and debug code outside of the hardware laboratory environment, making it an excellent, economical software development tool.

24.7 MPLAB ICE 2000 High-Performance In-Circuit Emulator

The MPLAB ICE 2000 In-Circuit Emulator is intended to provide the product development engineer with a complete microcontroller design tool set for PIC microcontrollers. Software control of the MPLAB ICE 2000 In-Circuit Emulator is advanced by the MPLAB Integrated Development Environment, which allows editing, building, downloading and source debugging from a single environment.

The MPLAB ICE 2000 is a full-featured emulator system with enhanced trace, trigger and data monitoring features. Interchangeable processor modules allow the system to be easily reconfigured for emulation of different processors. The architecture of the MPLAB ICE 2000 In-Circuit Emulator allows expansion to support new PIC microcontrollers.

The MPLAB ICE 2000 In-Circuit Emulator system has been designed as a real-time emulation system with advanced features that are typically found on more expensive development tools. The PC platform and Microsoft[®] Windows[®] 32-bit operating system were chosen to best make these features available in a simple, unified application.

24.8 MPLAB REAL ICE In-Circuit Emulator System

MPLAB REAL ICE In-Circuit Emulator System is Microchip's next generation high-speed emulator for Microchip Flash DSC[®] and MCU devices. It debugs and programs PIC[®] and dsPIC[®] Flash microcontrollers with the easy-to-use, powerful graphical user interface of the MPLAB Integrated Development Environment (IDE), included with each kit.

The MPLAB REAL ICE probe is connected to the design engineer's PC using a high-speed USB 2.0 interface and is connected to the target with either a connector compatible with the popular MPLAB ICD 2 system (RJ11) or with the new high speed, noise tolerant, lowvoltage differential signal (LVDS) interconnection (CAT5).

MPLAB REAL ICE is field upgradeable through future firmware downloads in MPLAB IDE. In upcoming releases of MPLAB IDE, new devices will be supported, and new features will be added, such as software breakpoints and assembly code trace. MPLAB REAL ICE offers significant advantages over competitive emulators including low-cost, full-speed emulation, real-time variable watches, trace analysis, complex breakpoints, a ruggedized probe interface and long (up to three meters) interconnection cables.

24.9 MPLAB ICD 2 In-Circuit Debugger

Microchip's In-Circuit Debugger, MPLAB ICD 2, is a powerful, low-cost, run-time development tool, connecting to the host PC via an RS-232 or high-speed USB interface. This tool is based on the Flash PIC MCUs and can be used to develop for these and other PIC MCUs and dsPIC DSCs. The MPLAB ICD 2 utilizes the in-circuit debugging capability built into the Flash devices. This feature, along with Microchip's In-Circuit Serial Programming[™] (ICSP[™]) protocol, offers costeffective, in-circuit Flash debugging from the graphical user interface of the MPLAB Integrated Development Environment. This enables a designer to develop and debug source code by setting breakpoints, single stepping and watching variables, and CPU status and peripheral registers. Running at full speed enables testing hardware and applications in real time. MPLAB ICD 2 also serves as a development programmer for selected PIC devices.

24.10 MPLAB PM3 Device Programmer

The MPLAB PM3 Device Programmer is a universal, CE compliant device programmer with programmable voltage verification at VDDMIN and VDDMAX for maximum reliability. It features a large LCD display (128 x 64) for menus and error messages and a modular, detachable socket assembly to support various package types. The ICSP[™] cable assembly is included as a standard item. In Stand-Alone mode, the MPLAB PM3 Device Programmer can read, verify and program PIC devices without a PC connection. It can also set code protection in this mode. The MPLAB PM3 connects to the host PC via an RS-232 or USB cable. The MPLAB PM3 has high-speed communications and optimized algorithms for quick programming of large memory devices and incorporates an SD/MMC card for file storage and secure data applications.

24.11 PICSTART Plus Development Programmer

The PICSTART Plus Development Programmer is an easy-to-use, low-cost, prototype programmer. It connects to the PC via a COM (RS-232) port. MPLAB Integrated Development Environment software makes using the programmer simple and efficient. The PICSTART Plus Development Programmer supports most PIC devices in DIP packages up to 40 pins. Larger pin count devices, such as the PIC16C92X and PIC17C76X, may be supported with an adapter socket. The PICSTART Plus Development Programmer is CE compliant.

24.12 PICkit 2 Development Programmer

The PICkit[™] 2 Development Programmer is a low-cost programmer and selected Flash device debugger with an easy-to-use interface for programming many of Microchip's baseline, mid-range and PIC18F families of Flash memory microcontrollers. The PICkit 2 Starter Kit includes a prototyping development board, twelve sequential lessons, software and HI-TECH's PICC[™] Lite C compiler, and is designed to help get up to speed quickly using PIC[®] microcontrollers. The kit provides everything needed to program, evaluate and develop applications using Microchip's powerful, mid-range Flash memory family of microcontrollers.

24.13 Demonstration, Development and Evaluation Boards

A wide variety of demonstration, development and evaluation boards for various PIC MCUs and dsPIC DSCs allows quick application development on fully functional systems. Most boards include prototyping areas for adding custom circuitry and provide application firmware and source code for examination and modification.

The boards support a variety of features, including LEDs, temperature sensors, switches, speakers, RS-232 interfaces, LCD displays, potentiometers and additional EEPROM memory.

The demonstration and development boards can be used in teaching environments, for prototyping custom circuits and for learning about various microcontroller applications.

In addition to the PICDEM[™] and dsPICDEM[™] demonstration/development board series of circuits, Microchip has a line of evaluation kits and demonstration software for analog filter design, KEELOQ[®] security ICs, CAN, IrDA[®], PowerSmart[®] battery management, SEEVAL[®] evaluation system, Sigma-Delta ADC, flow rate sensing, plus many more.

Check the Microchip web page (www.microchip.com) and the latest *"Product Selector Guide"* (DS00148) for the complete list of demonstration, development and evaluation kits.

25.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings^(†)

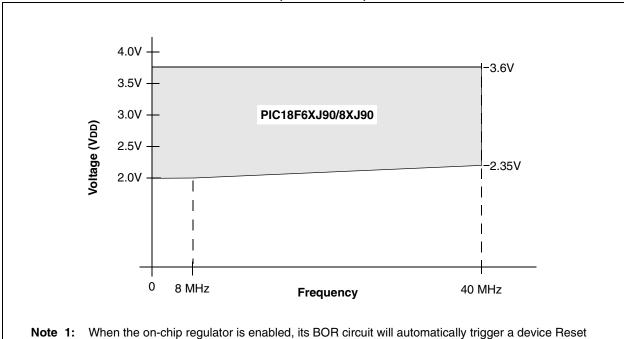
Ambient temperature under bias40°C to +100°C
Storage temperature65°C to +150°C
Voltage on any digital only I/O pin or MCLR with respect to Vss (except VDD)
Voltage on any combined digital and analog pin with respect to Vss (except VDD and MCLR)0.3V to (VDD + 0.3V)
Voltage on VDDCORE with respect to Vss0.3V to 2.75V
Voltage on VDD with respect to Vss
Total power dissipation (Note 1)1.0W
Maximum current out of Vss pin
Maximum current into VDD pin250 mA
Maximum output current sunk by PORTA<7:6> and any PORTB and PORTC I/O pins
Maximum output current sunk by any PORTD, PORTE and PORTJ I/O pins
Maximum output current sunk by PORTA<5:0> and any PORTF, PORTG and PORTH I/O pins
Maximum output current sourced by PORTA<7:6> and any PORTB and PORTC I/O pins
Maximum output current sourced by any PORTD, PORTE and PORTJ I/O pins
Maximum output current sourced by PORTA<5:0> and any PORTF, PORTG and PORTH I/O pins
Maximum current sunk by all ports combined
Maximum output current sunk by PORTA<5:0> and any PORTF, PORTG and PORTH I/O pins

Note 1: Power dissipation is calculated as follows:

 $\mathsf{Pdis} = \mathsf{VDD} \ \mathsf{x} \ \{\mathsf{IDD} - \sum \mathsf{IOH}\} + \sum \{(\mathsf{VDD} - \mathsf{VOH}) \ \mathsf{x} \ \mathsf{IOH}\} + \sum (\mathsf{VOL} \ \mathsf{x} \ \mathsf{IOL})$

† NOTICE: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

FIGURE 25-1: PIC18F85J90 FAMILY VOLTAGE-FREQUENCY GRAPH, REGULATOR ENABLED (INDUSTRIAL)⁽¹⁾



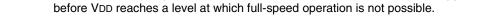
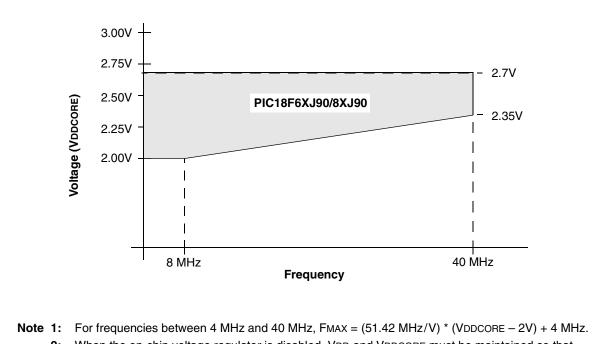


FIGURE 25-2: PIC18F85J90 FAMILY VOLTAGE-FREQUENCY GRAPH, REGULATOR DISABLED (INDUSTRIAL)^(1,2)



When the on-chip voltage regulator is disabled, VDD and VDDCORE must be maintained so that VDDCORE ≤ VDD ≤ 3.6V.

25.1	DC Characteristics:	Supply Voltage
		PIC18F85J90 Family (Industrial)

	85J90 Fam ustrial)	hily	Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial					
Param No.	Symbol	Characteristic	Min	Тур	Max	Units	Conditions	
D001	Vdd	Supply Voltage	VDDCORE 2.0	_	3.6 3.6	V V	ENVREG tied to Vss ENVREG tied to Vpp	
D001B	VDDCORE	External Supply for Microcontroller Core	2.0	—	2.70	V	ENVREG tied to Vss	
D001C	AVdd	Analog Supply Voltage	Vdd - 0.3	_	VDD + 0.3	V		
D001D	AVss	Analog Ground Potential	Vss - 0.3	_	Vss + 0.3	V		
D002	Vdr	RAM Data Retention Voltage ⁽¹⁾	1.5	—	—	V		
D003	VPOR	VDD Start Voltage to ensure internal Power-on Reset signal		_	0.7	V	See Section 4.3 "Power-on Reset (POR)" for details	
D004	SVDD	VDD Rise Rate to ensure internal Power-on Reset signal	0.05	—	—	V/ms	See Section 4.3 "Power-on Reset (POR)" for details	
D005	VBOR	Brown-out Reset Voltage	—	1.8	—	V		

Note 1: This is the limit to which VDD can be lowered in Sleep mode, or during a device Reset, without losing RAM data.

PIC18F8 (Indus	5 J90 Family strial)	Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial						
Param No.	Device	Тур	Max	Units	Conditions			
	Power-Down Current (IPD)	1)						
	All devices	0.2	0.9	μA	-40°C	VDD = 2.0V,		
		0.1	0.9	μA	+25°C	VDDCORE = 2.0V		
		2.4	5	μA	+85°C	(Sleep mode) ⁽⁴⁾		
	All devices	0.5	0.9	μA	-40°C	VDD = 2.5V,		
		0.1	0.9	μA	+25°C	VDDCORE = 2.5V		
		2.7	5	μA	+85°C	(Sleep mode) ⁽⁴⁾		
	All devices	2.7	6	μA	-40°C			
		3.5	6	μA	+25°C	VDD = 3.3V (Sleep mode) ⁽⁵⁾		
		6.7	12	μA	+85°C			

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD or Vss, and all features that add delta current disabled (such as WDT, Timer1 oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

- **3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4: Voltage regulator disabled (ENVREG tied to Vss).
- **5:** Voltage regulator enabled (ENVREG tied to VDD).
- 6: Resistor ladder current is not included.
- 7: Connecting an actual display will increase the current consumption depending on the size of the LCD.

	5J90 Family strial)	Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial							
Param No.	Device	Тур	Max	S					
	Supply Current (IDD) ⁽²⁾								
	All devices	6.5	16	μA	-40°C				
		7	16	μA	+25°C	VDD = 2.0V, $VDDCORE = 2.0V(4)$			
		9.5	20	μA	+85°C				
	All devices	10	18	μA	-40°C		Fosc = 31 kHz		
		10.5	18	μA	+25°C	VDD = 2.5V, $VDDCORE = 2.5V(4)$	(INTRC_RUN mode,		
		12.5	24	μA	+85°C		internal oscillator source		
	All devices	41	100	μA	-40°C				
		52	100	μA	+25°C	VDD = 3.3V ⁽⁵⁾			
		71	110	μA	+85°C				
	All devices	359	750	μA	-40°C		Fosc = 1 MHz (INTOSC_RUN mode,		
		387	750	μA	+25°C	VDD = 2.0V, $VDDCORE = 2.0V(4)$			
		407	840	μA	+85°C	VDDCORE = 2.000			
	All devices	438	850	μA	-40°C				
		470	850	μA	+25°C	VDD = 2.5V, VDDCORE = 2.5V ⁽⁴⁾			
		491	910	μA	+85°C		internal oscillator source		
	All devices	486	900	μA	-40°C				
		526	900	μA	+25°C	VDD = 3.3V ⁽⁵⁾			
		564	990	μA	+85°C				
	All devices	0.76	1.45	mA	-40°C				
		0.84	1.45	mA	+25°C	VDD = 2.0V, $VDDCORE = 2.0V(4)$			
		0.9	1.6	mA	+85°C	VDD00112 - 2.0V			
	All devices	1.1	1.63	mA	-40°C		Fosc = 4 MHz		
		1.18	1.63	mA	+25°C	VDD = 2.5V, VDDCORE = 2.5V ⁽⁴⁾	(INTOSC_RUN mode,		
		1.24	1.75	mA	+85°C	70000112 - 2.0VV	internal oscillator source		
	All devices	1.25	1.86	mA	-40°C				
		1.29	1.86	mA	+25°C	VDD = 3.3V ⁽⁵⁾			
		1.37	1.94	mA	+85°C				

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD or Vss, and all features that add delta current disabled (such as WDT, Timer1 oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

- **3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4: Voltage regulator disabled (ENVREG tied to Vss).
- 5: Voltage regulator enabled (ENVREG tied to VDD).
- 6: Resistor ladder current is not included.
- 7: Connecting an actual display will increase the current consumption depending on the size of the LCD.

25.2 DC Characteristics: Power-Down and Supply Current PIC18F85J90 Family (Industrial) (Continued)

PIC18F85J90 Family (Industrial)		Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial							
Param No.	Device	Тур	Max	Units		s			
	Supply Current (IDD) ⁽²⁾								
	All devices	2.4	8	μA	-40°C				
		2.5	8	μA	+25°C	VDD = 2.0V, $VDDCORE = 2.0V(4)$			
		4.8	12	μA	+85°C	VBBCONE = 2.0V			
	All devices	3.2	9	μA	-40°C		Fosc = 31 kHz		
		3.2	9	μA	+25°C	VDD = 2.5V, VDDCORE = 2.5V ⁽⁴⁾	(INTRC_IDLE mode,		
		6	14	μA	+85°C	VDD00112 - 2.3V	internal oscillator source		
	All devices	62	82	μA	-40°C				
		42	82	μA	+25°C	VDD = 3.3V ⁽⁵⁾			
		59	97	μA	+85°C				
	All devices	251	570	μA	-40°C		Fosc = 1 MHz (INTOSC_IDLE mode,		
		264	570	μA	+25°C	VDD = 2.0V, $VDDCORE = 2.0V(4)$			
		272	590	μA	+85°C	VDDCORE = 2.0V**			
	All devices	284	610	μA	-40°C				
		284	610	μA	+25°C	VDD = 2.5V, VDDCORE = 2.5V ⁽⁴⁾			
		293	650	μA	+85°C	VDD00112 - 2.3V	internal oscillator sourc		
	All devices	295	710	μA	-40°C				
		323	710	μA	+25°C	VDD = 3.3V ⁽⁵⁾			
		392	790	μA	+85°C				
	All devices	368	760	μA	-40°C				
		362	760	μA	+25°C	VDD = 2.0V, $VDDCORE = 2.0V(4)$			
		370	800	μA	+85°C				
	All devices	400	850	μA	-40°C		Fosc = 4 MHz		
		410	850	μA	+25°C	VDD = 2.5V, VDDCORE = 2.5V ⁽⁴⁾	(INTOSC_IDLE mode		
		418	900	μA	+85°C		internal oscillator sourc		
	All devices	460	950	μA	-40°C				
		462	950	μA	+25°C	VDD = 3.3V ⁽⁵⁾			
		486	1000	μA	+85°C				

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD or VSS, and all features that add delta current disabled (such as WDT, Timer1 oscillator, BOR, etc.).

- 2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.
 - The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

- MCLR = VDD; WDT enabled/disabled as specified.
- **3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4: Voltage regulator disabled (ENVREG tied to Vss).
- 5: Voltage regulator enabled (ENVREG tied to VDD).
- 6: Resistor ladder current is not included.
- 7: Connecting an actual display will increase the current consumption depending on the size of the LCD.

	5J90 Family strial)	Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial								
Param No.	Device	Тур	Max	Units		Conditions				
	Supply Current (IDD) ⁽²⁾									
	All devices	165	490	μA	-40°C					
		180	490	μA	+25°C	VDD = 2.0V, $VDDCORE = 2.0V(4)$				
		200	490	μA	+85°C	VBBOONE = 2.0V				
	All devices	256	670	μA	-40°C		Fosc = 1 MHz			
		260	670	μA	+25°C	VDD = 2.5V, $VDDCORE = 2.5V^{(4)}$	(PRI_RUN mode,			
		280	670	μA	+85°C		EC oscillator)			
	All devices	460	850	μA	-40°C					
		456	850	μA	+25°C	VDD = 3.3V ⁽⁵⁾				
		482	850	μA	+85°C					
	All devices	0.632	2.2	mA	-40°C		Fosc = 4 MHz			
		0.681	2.2	mA	+25°C	VDD = 2.0V, $VDDCORE = 2.0V(4)$				
		0.738	2.2	mA	+85°C					
	All devices	0.912	2.5	mA	-40°C					
		1.04	2.5	mA	+25°C	VDD = 2.5V, $VDDCORE = 2.5V(4)$	(PRI_RUN mode,			
		1.04	2.5	mA	+85°C		EC oscillator)			
	All devices	1.32	3.0	mA	-40°C					
		1.32	3.0	mA	+25°C	VDD = 3.3V ⁽⁵⁾				
		1.41	3.0	mA	+85°C					
	All devices	7.47	14	mA	-40°C					
		5.81	14	mA	+25°C	VDD = 2.5V, $VDDCORE = 2.5V(4)$	E			
		6.32	13	mA	+85°C		Fosc = 40 MHz (PRI_RUN mode,			
	All devices	8.84	18	mA	-40°C		EC oscillator)			
		8.66	18	mA	+25°C	VDD = 3.3V ⁽⁵⁾	,			
		7.97	16	mA	+85°C					

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD or VSS, and all features that add delta current disabled (such as WDT, Timer1 oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

- **3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4: Voltage regulator disabled (ENVREG tied to Vss).
- 5: Voltage regulator enabled (ENVREG tied to VDD).
- 6: Resistor ladder current is not included.
- 7: Connecting an actual display will increase the current consumption depending on the size of the LCD.

PIC18F85J90 Family (Industrial)		Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial										
Param No.	Device	Тур	Max	Units	Conditions							
	Supply Current (IDD) ⁽²⁾											
	All devices	2.8	3.8	mA	-40°C		Fosc = 4 MHz,					
		3.02	3.8	mA	+25°C	VDD = 2.0V, $VDDCORE = 2.0V^{(4)}$	16 MHz internal					
		3.01	4.5	mA	+85°C	$\forall DDCORE = 2.0V^{1/2}$	(PRI_RUN mode, HSPLL oscillator)					
	All devices	4.5	5.4	mA	-40°C		Fosc = 4 MHz,					
		4.8	5.6	mA	+25°C	VDD = 2.5V, $VDDCORE = 2.5V(4)$	16 MHz internal					
		4.54	5.6	mA	+85°C	$\forall DDCORE = 2.5V^{(1)}$	(PRI_RUN mode, HSPLL oscillator)					
	All devices	5.72	6.7	mA	-40°C		Fosc = 4 MHz,					
		5.55	6.5	mA	+25°C	VDD = 3.3V ⁽⁵⁾	16 MHz internal					
		5.3	6.5	mA	+85°C		(PRI_RUN mode, HSPLL oscillator)					
	All devices	7.4	8.5	mA	-40°C		Fosc = 10 MHz,					
		7.23	8.5	mA	+25°C	VDD = 2.5V,	40 MHz internal					
		6.55	7.5	mA	+85°C	VDDCORE = 2.5V ⁽⁴⁾	(PRI_RUN mode, HSPLL oscillator)					
	All devices	9.74	11.6	mA	-40°C		Fosc = 10 MHz,					
		9.43	11.6	mA	+25°C	VDD = 3.3V ⁽⁵⁾	40 MHz internal					
		8.89	10.5	mA	+85°C		(PRI_RUN mode, HSPLL oscillator)					

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD or VSS, and all features that add delta current disabled (such as WDT, Timer1 oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

- **3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4: Voltage regulator disabled (ENVREG tied to Vss).
- **5:** Voltage regulator enabled (ENVREG tied to VDD).
- 6: Resistor ladder current is not included.
- 7: Connecting an actual display will increase the current consumption depending on the size of the LCD.

	5J90 Family strial)	Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial								
Param No.	Device	Тур	Max	Units		Conditions	;			
	Supply Current (IDD) ⁽²⁾									
	All devices	50	120	μA	-40°C					
		51	120	μA	+25°C	$VDD = 2.0V,$ $VDDCORE = 2.0V^{(4)}$				
		54	130	μA	+85°C	VDD00112 - 2.0V				
	All devices	223	480	μA	-40°C)/ 0.5\/	Fosc = 1 MHz			
		133	300	μA	+25°C	$VDD = 2.5V,$ $VDDCORE = 2.5V^{(4)}$	(PRI_IDLE mode,			
		110	270	μA	+85°C	VBBOONE - E.OV	EC oscillator)			
	All devices	307	550	μA	-40°C					
		254	500	μA	+25°C	VDD = 3.3V ⁽⁵⁾				
		194	460	μA	+85°C					
	All devices	307	850	μA	-40°C		Fosc = 4 MHz			
		200	850	μA	+25°C	$VDD = 2.0V,$ $VDDCORE = 2.0V^{(4)}$				
		202	800	μA	+85°C					
	All devices	483	950	μA	-40°C					
		318	950	μA	+25°C	VDD = 2.5V, VDDCORE = 2.5V ⁽⁴⁾	(PRI_IDLE mode,			
		343	900	μA	+85°C		EC oscillator)			
	All devices	524	1.3	mA	-40°C					
		474	1.2	mA	+25°C	VDD = 3.3V ⁽⁵⁾				
		468	1.2	mA	+85°C					
	All devices	2.38	8	mA	-40°C					
		2.04	8	mA	+25°C	$VDD = 2.5V,$ $VDDCORE = 2.5V^{(4)}$				
		2.52	9	mA	+85°C		Fosc = 40 MHz (PRI_IDLE mode,			
	All devices	3.02	10	mA	-40°C		EC oscillator)			
		2.99	10	mA	+25°C	VDD = 3.3V ⁽⁵⁾	,			
		4.23	11	mA	+85°C					

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD or VSS, and all features that add delta current disabled (such as WDT, Timer1 oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

- **3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4: Voltage regulator disabled (ENVREG tied to Vss).
- 5: Voltage regulator enabled (ENVREG tied to VDD).
- 6: Resistor ladder current is not included.
- 7: Connecting an actual display will increase the current consumption depending on the size of the LCD.

PIC18F85J90 Family (Industrial)		Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial										
Param No.	Device	Тур	Max	Units	Conditions							
Supply Current (IDD) ⁽²⁾												
	All devices	10.5	22	μA	-40°C							
		13.4	28	μA	+25°C	VDD = 2.0V, $VDDCORE = 2.0V(4)$						
		17.6	40	μA	+85°C	VBB00112 - 2.0V						
	All devices	13.2	30	μA	-40°C		Fosc = 32 kHz ⁽³⁾					
		16.2	35	μA	+25°C	$VDD = 2.5V,$ $VDDCORE = 2.5V^{(4)}$	(SEC_RUN mode, Timer1 as clock)					
		20.7	50	μA	+85°C	V B B B B B B B B B B B B B B B B B B B						
	All devices	39	120	μA	-40°C							
		58	150	μA	+25°C	VDD = 3.3V ⁽⁵⁾						
		75	190	μA	+85°C							
	All devices	5.7	15	μA	-40°C							
		8.9	20	μA	+25°C	VDD = 2.0V, $VDDCORE = 2.0V(4)$						
		12.8	26	μA	+85°C	VBBOONE - 2.0V						
	All devices	6.6	17	μA	-40°C		Fosc = 32 kHz ⁽³⁾					
		9.7	24	μA	+25°C	VDD = 2.5V, $VDDCORE = 2.5V(4)$	(SEC_IDLE mode,					
		13.7	30	μA	+85°C		Timer1 as clock)					
	All devices	39	115	μA	-40°C							
		52.8	145	μA	+25°C	VDD = 3.3V ⁽⁵⁾						
		72.7	185	μA	+85°C							

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD or Vss, and all features that add delta current disabled (such as WDT, Timer1 oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

- **3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4: Voltage regulator disabled (ENVREG tied to Vss).
- 5: Voltage regulator enabled (ENVREG tied to VDD).
- 6: Resistor ladder current is not included.
- 7: Connecting an actual display will increase the current consumption depending on the size of the LCD.

25.2 DC Characteristics: Power-Down and Supply Current PIC18F85J90 Family (Industrial) (Continued)

PIC18F8 (Indus	•	Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial									
Param No.	Device	Тур	Мах	Units		Condition	S				
	Module Differential Curren	ts (Δ Iwdt, Δ	ILCD, ∆I	oscв, ∆I	AD)						
D022	Watchdog Timer	1.6	4	μA	-40°C	VDD = 2.0V,					
(∆Iwdt)		1.7	4	μA	+25°C	VDD = 2.0V, VDDCORE = 2.0V ⁽⁴⁾					
		1.6	4	μA	+85°C						
		2.5	5	μA	-40°C	VDD = 2.5V,					
		2.5	5	μA	+25°C	VDD = 2.5V, $VDDCORE = 2.5V^{(4)}$					
		2.3	5	μA	+85°C						
		3.8	6	μA	-40°C						
		2.6	6	μA	+25°C	VDD = 3.3V ⁽⁵⁾					
		2.4	6	μA	+85°C						
D024	LCD Module		5	μA	+25°C	VDD = 2.0V	Resistive Ladder CPEN = 0;				
$(\Delta ILCD)$		2.7 ^(6,7)	5	μA	+25°C	VDD = 2.5V					
		3.5 ^(6,7)	7	μΑ	+25°C	VDD = 3.0V	CKSEL<1:0> = 00; CS<1:0> = 10;				
		16 ⁽⁷⁾	25	μA	+25°C	VDD = 2.0V	Charge Pump				
		17(7)	25	μA	+25°C	VDD = 2.5V	BIAS<2:0> = 111;				
		24 ⁽⁷⁾	40	μA	+25°C	VDD = 3.0V	CPEN = 1; CKSEL<1:0> = 11;				
D025	Timer1 Oscillator	6.6	12.5	μA	-40°C						
(Δ IOSCB)		7.9	12.5	μA	+25°C	VDD = 2.0V, $VDDCORE = 2.0V^{(4)}$	32 kHz on Timer1 ⁽³⁾				
		11.5	18.5	μA	+85°C	VDDCORE = 2.0V					
		7.2	12.5	μA	-40°C						
		8.1	12.5	μA	+25°C	VDD = 2.5V, $VDDCORE = 2.5V^{(4)}$	32 kHz on Timer1 ⁽³⁾				
		11.9	18.5	μA	+85°C	VDD00112 - 2.5V					
		7	12.5	μA	-40°C						
		9	12.5	μA	+25°C	VDD = 3.3V ⁽⁵⁾	32 kHz on Timer1 ⁽³⁾				
		11	18.5	μA	+85°C						
D026 (∆IAD)	A/D Converter	1	1.5	μA	-40°C to +85°C	VDD = 2.0V, $VDDCORE = 2.0V^{(4)}$					
		1	1.5	μA	-40°C to +85°C	VDD = 2.5V, $VDDCORE = 2.5V^{(4)}$	A/D on, not converting				
		1	1.5	μA	-40°C to +85°C	VDD = 3.3V ⁽⁵⁾	1				

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD or Vss, and all features that add delta current disabled (such as WDT, Timer1 oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

- 3: Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4: Voltage regulator disabled (ENVREG tied to Vss).
- 5: Voltage regulator enabled (ENVREG tied to VDD).
- 6: Resistor ladder current is not included.
- 7: Connecting an actual display will increase the current consumption depending on the size of the LCD.

25.3 DC Characteristics:PIC18F84J90 Family (Industrial)

DC CHA	ARACTE	RISTICS	Standard Operating Conditions (unless otherwise stated) Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial					
Param No.	Symbol	Characteristic	Min	Мах	Units	Conditions		
	Vi∟	Input Low Voltage						
		All I/O ports:						
D030		with TTL buffer	Vss	0.15 Vdd	V			
D031		with Schmitt Trigger buffer	Vss	0.2 Vdd	V			
D032		MCLR	Vss	0.2 Vdd	V			
D033		OSC1	Vss	0.3 Vdd	V	HS, HSPLL modes		
D033A		OSC1	Vss	0.2 Vdd	V	EC, ECPLL modes		
D034		T13CKI	Vss	0.3	V			
	Viн	Input High Voltage						
		I/O ports with analog functions:						
D040		with TTL buffer	0.25 VDD + 0.8V	Vdd	V	Vdd < 3.3V		
D041		with Schmitt Trigger buffer	0.8 Vdd	Vdd	V			
		Digital-only I/O ports:						
		with TTL buffer	0.25 VDD + 0.8V	5.5	V	Vdd < 3.3V		
			2.0	5.5	V	$3.3V \le VDD \le 3.6V$		
		with Schmitt Trigger buffer	0.8 Vdd	5.5	V			
D042		MCLR	0.8 Vdd	Vdd	V			
D043		OSC1	0.7 Vdd	Vdd	V	HS, HSPLL modes		
D043A		OSC1	0.8 Vdd	Vdd	V	EC, ECPLL modes		
D044		T13CKI	1.6	Vdd	V			
	lı∟	Input Leakage Current ⁽¹⁾						
D060		I/O ports	—	±1	μA	$Vss \le VPIN \le VDD,$ Pin at high-impedance		
D061		MCLR		±1	μA	$Vss \le VPIN \le VDD$		
D063		OSC1	—	±5	μA	$Vss \le VPIN \le VDD$		
	IPU	Weak Pull-up Current						
D070	IPURB	PORTB weak pull-up current	30	240	μA	VDD = 3.3V, VPIN = VSS		

Note 1: Negative current is defined as current sourced by the pin.

			Standard Operating Conditions (unless otherwise stated) Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial					
Param No.	Symbol	Characteristic	Min	Max	Units	Conditions		
	Vol	Output Low Voltage						
D080		I/O ports:						
		PORTA, PORTF, PORTG, PORTH	_	0.4	V	IOL = 2 mA, VDD = 3.3V, -40°C to +85°C		
		PORTD, PORTE, PORTJ	_	0.4	V	IOL = 3.4 mA, VDD = 3.3V, -40°C to +85°C		
		PORTB, PORTC	_	0.4	V	IOL = 3.4 mA, VDD = 3.3V, -40°C to +85°C		
D083		OSC2/CLKO (EC, ECPLL modes)	_	0.4	V	IOL = 1.6 mA, VDD = 3.3V, -40°C to +85°C		
	Vон	Output High Voltage ⁽¹⁾						
D090		I/O ports:			V			
		PORTA, PORTF, PORTG, PORTH	2.4	_	V	IOH = -2 mA, VDD = 3.3V, -40°С to +85°С		
		PORTD, PORTE, PORTJ	2.4	_	V	IOH = -2 mA, VDD = 3.3V, -40°С to +85°С		
		PORTB, PORTC	2.4	_	V	IOH = -2 mA, VDD = 3.3V, -40°С to +85°С		
D092		OSC2/CLKO (INTOSC, EC, ECPLL modes)	2.4	—	V	IOH = -1 mA, VDD = 3.3V, -40°С to +85°С		
		Capacitive Loading Specs on Output Pins						
D100 ⁽⁴⁾	COSC2	OSC2 pin	_	15	pF	In HS mode when external clock is used to drive OSC1		
D101	Сю	All I/O pins and OSC2	_	50	pF	To meet the AC Timing Specifications		
D102	Св	SCLx, SDAx	_	400	pF	I ² C [™] Specification		

25.3 DC Characteristics:PIC18F84J90 Family (Industrial) (Continued)

Note 1: Negative current is defined as current sourced by the pin.

DC CHA	DC CHARACTERISTICS			Standard Operating Conditions (unless otherwise stated) Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial				
Param No.	Sym	Sym Characteristic Min Typ† Max L		Units	Conditions			
		Program Flash Memory						
D130	Eр	Cell Endurance	1K	10K	—	E/W	-40°C to +85°C	
D131	Vpr	VDD for Read	VMIN	—	3.6	V	VMIN = Minimum operating voltage	
D132B	Vpew	VDD for Self-Timed Write	VMIN	—	3.6	V	Vмın = Minimum operating voltage	
D133A	Tiw	Self-Timed Write Cycle Time	_	2.8	—	ms		
D134	TRETD	Characteristic Retention	20	—	_	Year	Provided no other specifications are violated	
D135	IDDP	Supply Current during Programming	_	3	7	mA		
D1xxx	TWE	Writes per Erase Cycle	—		1		Per one physical word address	

TABLE 25-1:MEMORY PROGRAMMING REQUIREMENTS

† Data in "Typ" column is at 3.3V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

TABLE 25-2: COMPARATOR SPECIFICATIONS

Operating	Operating Conditions: $3.0V \le V_{DD} \le 3.6V$, $-40^{\circ}C \le T_A \le +85^{\circ}C$ (unless otherwise stated)								
Param No.	Sym	Characteristics	Min	Тур	Мах	Units	Comments		
D300	VIOFF	Input Offset Voltage		±5.0	±10	mV			
D301	VICM	Input Common Mode Voltage*	0	_	AVDD - 1.5	V			
D302	CMRR	Common Mode Rejection Ratio*	55	_	—	dB			
300	TRESP	Response Time ^{(1)*}	_	150	400	ns			
301	TMC2OV	Comparator Mode Change to Output Valid*	_	—	10	μs			

* These parameters are characterized but not tested.

Note 1: Response time measured with one comparator input at (AVDD - 1.5)/2, while the other input transitions from Vss to VDD.

TABLE 25-3: VOLTAGE REFERENCE SPECIFICATIONS

Operating	Operating Conditions: $3.0V \le VDD \le 3.6V$, $-40^{\circ}C \le TA \le +85^{\circ}C$ (unless otherwise stated)								
Param No.	Sym	Characteristics	Min	Тур	Max	Units	Comments		
D310	VRES	Resolution	VDD/24	_	VDD/32	LSb			
D311	VRAA	Absolute Accuracy	—	_	1/2	LSb			
D312	VRur	Unit Resistor Value (R)	—	2k	—	Ω			
310	TSET	Settling Time ⁽¹⁾	_		10	μs			

Note 1: Settling time measured while CVRR = 1 and CVR3:CVR0 transitions from '0000' to '1111'.

TABLE 25-4: INTERNAL VOLTAGE REGULATOR SPECIFICATIONS

Operatir	Operating Conditions: $-40^{\circ}C \le TA \le +85^{\circ}C$ (unless otherwise stated)								
Param No.SymCharacteristicsMinTypMaxUnitsCompared					Comments				
	VRGOUT	Regulator Output Voltage*		2.5		V			
	CEFC	External Filter Capacitor Value*	4.7	10		μF	Capacitor must be low ESR		

These parameters are characterized but not tested. Parameter numbers not yet assigned for these specifications.

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TABLE 25-5: INTERNAL LCD VOLTAGE REGULATOR SPECIFICATIONS

Operatir	Operating Conditions: $2.0V \le VDD \le 3.6V$, $-40^{\circ}C \le TA \le +85^{\circ}C$ (unless otherwise stated)										
Param No.	Sym	Characteristics	Min	Тур	Max	Units	Comments				
	C _{FLY}	Fly Back Capacitor	0.47	4.7		μF	Capacitor must be low ESR				
	V _{BIAS}	V _{PK-PK} between LCDBIAS0 &		3.40	3.6	V	BIAS2:BIAS0 = 111				
		LCDBIAS3		3.27		V	BIAS2:BIAS0 = 110				
				3.14		V	BIAS2:BIAS0 = 101				
				3.01		V	BIAS2:BIAS0 = 100				
				2.88		V	BIAS2:BIAS0 = 011				
				2.75		V	BIAS2:BIAS0 = 010				
				2.62		V	BIAS2:BIAS0 = 001				
				2.49		V	BIAS2:BIAS0 = 000				

25.4 AC (Timing) Characteristics

25.4.1 TIMING PARAMETER SYMBOLOGY

The timing parameter symbols have been created following one of the following formats:

1. TppS2ppS	8	3. Tcc:st	(I ² C specifications only)
2. TppS		4. Ts	(I ² C specifications only)
Т			
F	Frequency	Т	Time
Lowercase le	etters (pp) and their meanings:		
рр			
сс	CCP1	osc	OSC1
ck	CLKO	rd	RD
cs	CS	rw	RD or WR
di	SDI	sc	SCK
do	SDO	SS	SS
dt	Data in	tO	TOCKI
io	I/O port	t1	T13CKI
mc	MCLR	wr	WR
Uppercase le	etters and their meanings:		
S			
F	Fall	Р	Period
н	High	R	Rise
I	Invalid (High-impedance)	V	Valid
L	Low	Z	High-impedance
I ² C only			
AA	output access	High	High
BUF	Bus free	Low	Low
TCC:ST (I ² C s	specifications only)		
CC			
HD	Hold	SU	Setup
ST			
DAT	DATA input hold	STO	Stop condition
STA	Start condition		

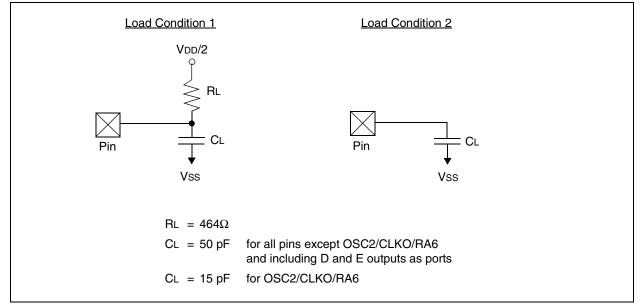
25.4.2 TIMING CONDITIONS

The temperature and voltages specified in Table 25-6 apply to all timing specifications unless otherwise noted. Figure 25-3 specifies the load conditions for the timing specifications.

TABLE 25-6: TEMPERATURE AND VOLTAGE SPECIFICATIONS – AC

	Standard Operating Conditions (unless otherwise stated)						
AC CHARACTERISTICS	Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial						
	Operating voltage VDD range as described in Section 25.1 and Section 25.3 .						

FIGURE 25-3: LOAD CONDITIONS FOR DEVICE TIMING SPECIFICATIONS



25.4.3 TIMING DIAGRAMS AND SPECIFICATIONS

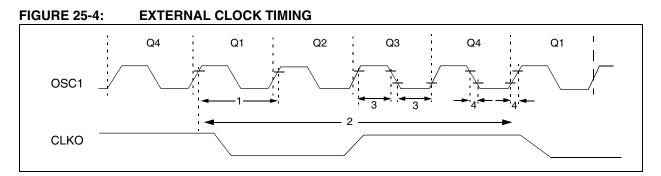


TABLE 25-7: EXTERNAL CLOCK TIMING REQUIREMENTS

Param. No.	Symbol	Characteristic	Min	Мах	Units	Conditions
1A	Fosc	External CLKI Frequency ⁽¹⁾	DC	40	MHz	ECPLL Oscillator mode
		Oscillator Frequency ⁽¹⁾	DC	40	MHz	HSPLL Oscillator mode
1	Tosc	External CLKI Period ⁽¹⁾	25	_	ns	EC Oscillator mode
		Oscillator Period ⁽¹⁾	25	250	ns	HS Oscillator mode
2	Тсү	Instruction Cycle Time ⁽¹⁾	100	_	ns	Tcy = 4/Fosc, Industrial
3	TosL, TosH	External Clock in (OSC1) High or Low Time	10	_	ns	EC Oscillator mode
4	TosR, TosF	External Clock in (OSC1) Rise or Fall Time	—	7.5	ns	EC Oscillator mode

Note 1: Instruction cycle period (TCY) equals four times the input oscillator time base period for all configurations except PLL. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min." values with an external clock applied to the OSC1/CLKI pin. When an external clock input is used, the "max." cycle time limit is "DC" (no clock) for all devices.

Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions
F10	Fosc	Oscillator Frequency Range	4		10	MHz	HS mode only
F11	Fsys	On-Chip VCO System Frequency	16		40	MHz	HS mode only
F12	t _{rc}	PLL Start-up Time (Lock Time)	_	_	2	ms	
F13	ΔCLK	CLKO Stability (Jitter)	-2	_	+2	%	

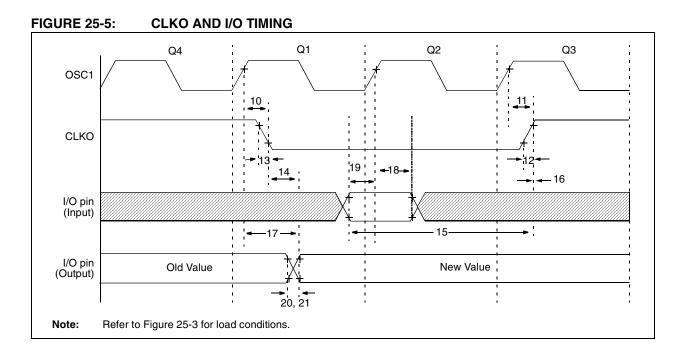
TABLE 25-8: PLL CLOCK TIMING SPECIFICATIONS (VDD = 2.15V TO 3.6V)

† Data in "Typ" column is at 3.3V, 25°C, unless otherwise stated. These parameters are for design guidance only and are not tested.

TABLE 25-9: INTERNAL RC ACCURACY (INTOSC AND INTRC SOURCES)

PIC18F85J90 Family (Industrial)			Standard Operating Conditions (unless otherwise stated) Deprating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial						
Param No.	Device		Тур	Max	Units	Condi	tions		
	INTOSC Accuracy @ Fre	q = 8 MHz	z, 4 MHz,	2 MHz, 1	MHz, 500) kHz, 250 kHz, 125 k	Hz, 31 kHz ⁽¹⁾		
	All Devices	-2	+/-1	2	%	+25°C	VDD = 2.7-3.3V		
		-5	_	5	%	-10°C to +85°C	VDD = 2.0-3.3V		
		-10	+/-1	10	%	-40°C to +85°C	VDD = 2.0-3.3V		
	INTRC Accuracy @ Freq	= 31 kHz	(1)						
	All Devices	26.562	—	35.938	kHz	-40°C to +85°C	VDD = 2.0-3.3V		

Note 1: The accuracy specification of the 31 kHz clock is determined by which source is providing it at a given time. When INTSRC (OSCTUNE<7>) is '1', use the INTOSC accuracy specification. When INTSRC is '0', use the INTRC accuracy specification.



Param No.	Symbol	Characteristic	Min	Тур	Мах	Units	Conditions
10	TosH2ckL	OSC1 ↑ to CLKO \downarrow	—	75	200	ns	(Note 1)
11	TosH2ckH	OSC1 ↑ to CLKO ↑	—	75	200	ns	(Note 1)
12	TCKR	CLKO Rise Time	—	15	30	ns	(Note 1)
13	ТскF	CLKO Fall Time	—	15	30	ns	(Note 1)
14	TCKL2IOV	CLKO \downarrow to Port Out Valid	—	_	0.5 Tcy + 20	ns	
15	TIOV2CKH	Port In Valid before CLKO \uparrow	0.25 TCY + 25		—	ns	
16	TckH2iol	Port In Hold after CLKO \uparrow	0			ns	
17	TosH2IoV	OSC1 \uparrow (Q1 cycle) to Port Out Valid	—	50	150	ns	
18	TosH2iol	OSC1 [↑] (Q2 cycle) to Port Input Invalid (I/O in hold time)	100	_	—	ns	
19	TioV2osH	Port Input Valid to OSC1 ↑ (I/O in setup time)	0	_	—	ns	
20	TIOR	Port Output Rise Time	—	_	6	ns	
21	TIOF	Port Output Fall Time	—	_	5	ns	
22†	TINP	INT pin High or Low Time	Тсү	_	—	ns	
23†	Trbp	RB7:RB4 Change INT High or Low Time	Тсү	—	—	ns	

TABLE 25-10: CLKO AND I/O TIMING REQUIREMENTS

† These parameters are asynchronous events not related to any internal clock edges.

Note 1: Measurements are taken in EC mode, where CLKO output is 4 x Tosc.

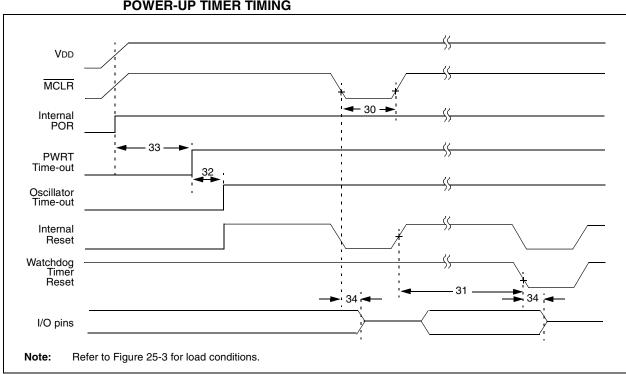


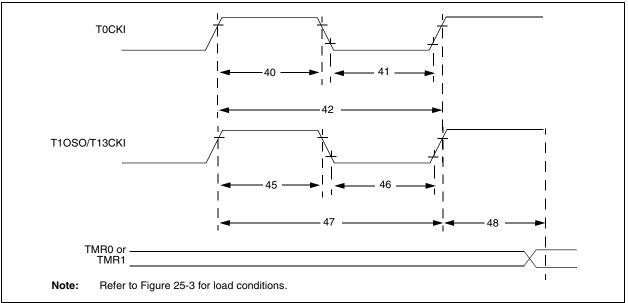
FIGURE 25-6: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND POWER-UP TIMER TIMING

TABLE 25-11: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER, POWER-UP TIMER AND BROWN-OUT RESET REQUIREMENTS

Param. No.	Symbol	Characteristic	Min	Тур	Мах	Units	Conditions
30	ТмсL	MCLR Pulse Width (low)	2 Tcy	10 Тсү	_		(Note 1)
31	Twdt	Watchdog Timer Time-out Period (no postscaler)	3.4	4.0	4.6	ms	
32	Tost	Oscillation Start-up Timer Period	1024 Tosc	_	1024 Tosc	_	Tosc = OSC1 period
33	TPWRT	Power-up Timer Period	45.8	65.5	85.2	ms	
34	Tioz	I/O High-Impedance from MCLR Low or Watchdog Timer Reset	—	2		μs	
38	TCSD	CPU Start-up Time	—	10	_	μs	
				200		μs	Voltage Regulator enabled and put to sleep
39	TIOBST	Time for INTOSC to Stabilize	—	1	_	μs	

Note 1: To ensure device Reset, MCLR must be low for at least 2 TCY or 400 µs, whichever is lower.





TADLL	ADLE 23-12. TIMERU AND TIMERT EXTERNAL CLOCK REQUIREMENTS							
Param No.	Symbol		Characteristic	:	Min	Max	Units	Conditions
40	T⊤0H	T0CKI High P	ulse Width	No prescaler	0.5 Tcy + 20	—	ns	
			V		10	—	ns	
41	T⊤0L	T0CKI Low P	ulse Width	No prescaler	0.5 Tcy + 20	—	ns	
				With prescaler	10	—	ns	
42	TT0P	T0CKI Period		No prescaler	Tcy + 10	—	ns	
				With prescaler	Greater of: 20 ns or (TCY + 40)/N	—	ns	N = prescale value (1, 2, 4,, 256)
45	T⊤1H	T13CKI High Time	Synchronous, no prescaler		0.5 Tcy + 20	—	ns	
			Synchronous, with prescaler		10	—	ns	
			Asynchronous		30	—	ns	
46	T⊤1L	T13CKI Low	Synchronous, n	o prescaler	0.5 TCY + 5	—	ns	
		Time	Synchronous, w	ith prescaler	10		ns	
			Asynchronous		30	—	ns	
47	TT1P	T13CKI Input Period	Synchronous		Greater of: 20 ns or (Tcy + 40)/N	—	ns	N = prescale value (1, 2, 4, 8)
			Asynchronous		60	—	ns	
	F⊤1	T13CKI Oscill	ator Input Freque	ency Range	DC	50	kHz	
48	TCKE2TMRI	Delay from Ex Timer Increme	kternal T13CKI C ent	lock Edge to	2 Tosc	7 Tosc	—	

FIGURE 25-8: CAPTURE/COMPARE/PWM TIMINGS (CCP1, CCP2 MODULES)

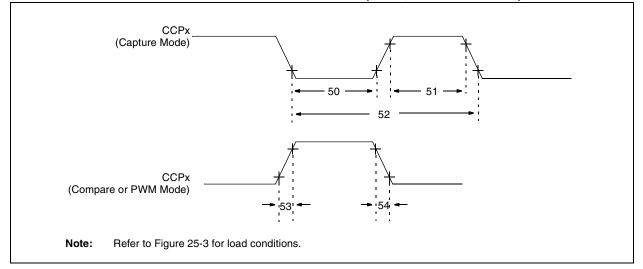


TABLE 25-13: CAPTURE/COMPARE/PWM REQUIREMENTS (CCP1, CCP2 MODULES)

Param No.	Symbol	с	haracteristic	Min	Max	Units	Conditions
50	TccL	CCPx Input Low	No prescaler	0.5 TCY + 20		ns	
	Tir	Time	With prescaler	10		ns	
51	TCCH	CCPx Input	No prescaler	0.5 TCY + 20		ns	
		High Time	With prescaler	10		ns	
52	TCCP	CCPx Input Perio	od	<u>3 Tcy + 40</u> N	_	ns	N = prescale value (1, 4 or 16)
53	TccR	CCPx Output Fal	ll Time	—	25	ns	
54	TCCF	CCPx Output Fal	ll Time	_	25	ns	

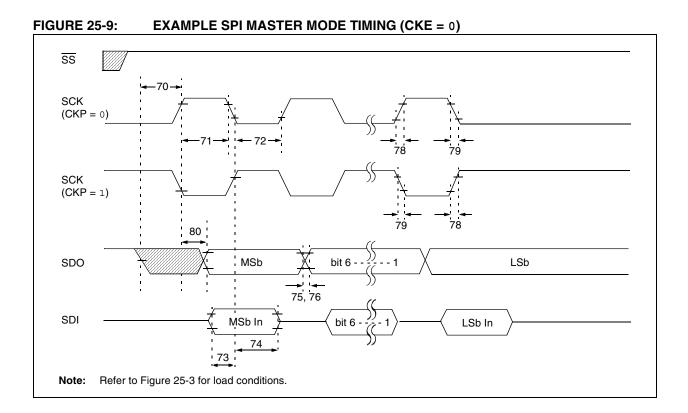


TABLE 25-14: EXAMPLE SPI MODE REQUIREMENTS (MASTER MODE, CKE = 0)

Param No.	Symbol	Characteristic		Min	Max	Units	Conditions
70	TssL2scH, TssL2scL	$\overline{SS} \downarrow$ to SCK \downarrow or SCK \uparrow Input		Тсү	—	ns	
71	TscH	SCK Input High Time	Continuous	1.25 TCY + 30	_	ns	
71A		(Slave mode)	Single Byte	40	_	ns	(Note 1)
72	TscL	SCK Input Low Time	Continuous	1.25 TCY + 30		ns	
72A		(Slave mode)	Single Byte	40	_	ns	(Note 1)
73	TDIV2SCH, TDIV2SCL	Setup Time of SDI Data Input to SCK Edge		100	—	ns	
73A	Тв2в	Last Clock Edge of Byte 1 to the of Byte 2	he 1st Clock Edge	1.5 TCY + 40		ns	(Note 2)
74	TscH2diL, TscL2diL	Hold Time of SDI Data Input to	o SCK Edge	100		ns	
75	TDOR	SDO Data Output Rise Time		—	25	ns	
76	TDOF	SDO Data Output Fall Time		—	25	ns	
78	TscR	SCK Output Rise Time (Master mode)		—	25	ns	
79	TscF	SCK Output Fall Time (Master mode)		—	25	ns	
80	TscH2doV, TscL2doV	SDO Data Output Valid after S	SCK Edge	—	50	ns	

Note 1: Requires the use of Parameter #73A.

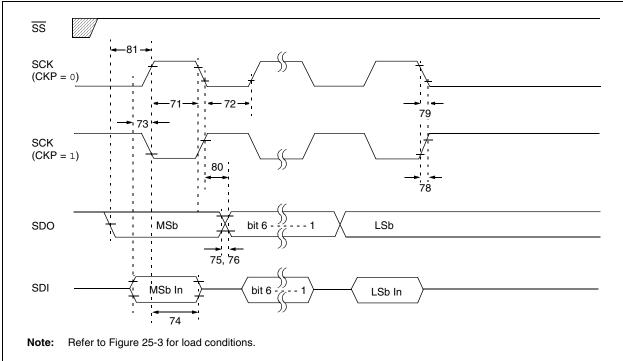


FIGURE 25-10: EXAMPLE SPI MASTER MODE TIMING (CKE = 1)

TABLE 25-15: EXAMPLE SPI MODE REQUIREMENTS (MASTER MODE, CKE = 1)

Param. No.	Symbol	Characteristic		Min	Max	Units	Conditions
71	TscH	SCK Input High Time	Continuous	1.25 Tcy + 30	—	ns	
71A		(Slave mode)	Single Byte	40		ns	(Note 1)
72	TscL	SCK Input Low Time	Continuous	1.25 TCY + 30	_	ns	
72A		(Slave mode)	Single Byte	40		ns	(Note 1)
73	TDIV2SCH, TDIV2SCL	Setup Time of SDI Data Input	to SCK Edge	100	—	ns	
73A	Тв2в	Last Clock Edge of Byte 1 to the 1st Clock Edge of Byte 2		1.5 TCY + 40	_	ns	(Note 2)
74	TscH2diL, TscL2diL	Hold Time of SDI Data Input t	o SCK Edge	100	_	ns	
75	TDOR	SDO Data Output Rise Time		—	25	ns	
76	TDOF	SDO Data Output Fall Time			25	ns	
78	TscR	SCK Output Rise Time (Maste	er mode)	—	25	ns	
79	TscF	SCK Output Fall Time (Maste	r mode)		25	ns	
80	TscH2doV, TscL2doV	SDO Data Output Valid after SCK Edge		—	50	ns	
81	TDOV2SCH, TDOV2SCL	SDO Data Output Setup to SO	CK Edge	Тсү	—	ns	

Note 1: Requires the use of Parameter #73A.

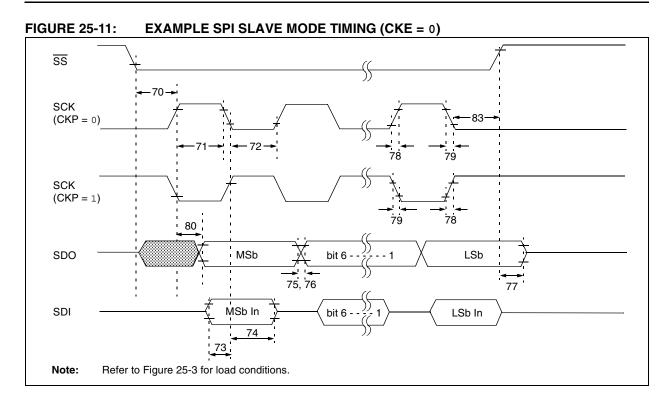


TABLE 25-16: EXAMPLE SPI MODE REQUIREMENTS (SLAVE MODE TIMING, CKE = 0)

Param No.	Symbol	Characteristic		Min	Max	Units	Conditions
70	TssL2scH, TssL2scL	$\overline{\mathrm{SS}}\downarrow$ to SCK \downarrow or SCK \uparrow Input	SCK ↑ Input		_	ns	
70A	TssL2WB	SS to write to SSPBUF		3 TCY	_	ns	
71	TSCH	SCK Input High Time (Slave mode)	Continuous	1.25 TCY + 30		ns	
71A			Single Byte	40		ns	(Note 1)
72	TscL	SCK Input Low Time	Continuous	1.25 TCY + 30	_	ns	
72A		(Slave mode)	Single Byte	40	_	ns	(Note 1)
73	TDIV2SCH, TDIV2SCL	Setup Time of SDI Data Input to SCK Edge		100	_	ns	
73A	Тв2в	Last Clock Edge of Byte 1 to the First Clo	ck Edge of Byte 2	1.5 Tcy + 40		ns	(Note 2)
74	TscH2DIL, TscL2DIL	Hold Time of SDI Data Input to SCK Ec	lge	100	_	ns	
75	TDOR	SDO Data Output Rise Time			25	ns	
76	TDOF	SDO Data Output Fall Time		_	25	ns	
77	TssH2doZ	SS ↑ to SDO Output High-impedance		10	50	ns	
78	TscR	SCK Output Rise Time (Master mode)		—	25	ns	
79	TscF	SCK Output Fall Time (Master mode)		—	25	ns	
80	TSCH2DOV, TSCL2DOV	SDO Data Output Valid after SCK Edge		—	50	ns	
83	TscH2ssH, TscL2ssH	SS ↑ after SCK Edge		1.5 TCY + 40		ns	

Note 1: Requires the use of Parameter #73A.

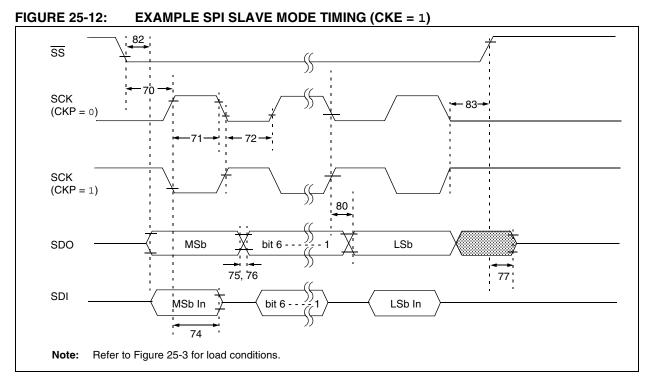
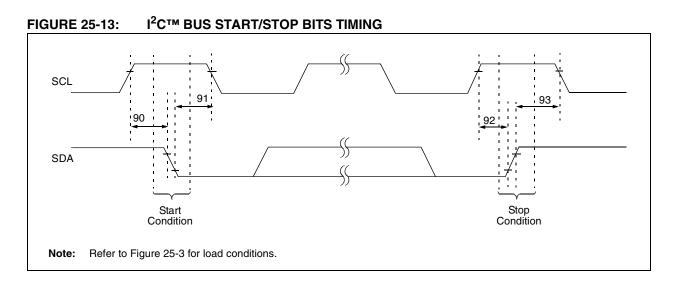


TABLE 25-17 F	EXAMPLE SPI SLAVE MODE REQUIREMENTS (CKE = 1)

Param No.	Symbol	Characteristic		Min	Max	Units	Conditions
70	TssL2scH, TssL2scL	$\overline{SS} \downarrow$ to SCK \downarrow or SCK \uparrow Input		3 Тсү	—	ns	
70A	TssL2WB	SS to write to SSPBUF		3 TCY	_	ns	
71	TscH	SCK Input High Time	Continuous	1.25 TCY + 30	—	ns	
71A		(Slave mode)	Single Byte	40	—	ns	(Note 1)
72	TscL	SCK Input Low Time	Continuous	1.25 Tcy + 30	—	ns	
72A		(Slave mode)	Single Byte	40	—	ns	(Note 1)
73A	Тв2в	Last Clock Edge of Byte 1 to the First	Clock Edge of Byte 2	1.5 TCY + 40	—	ns	(Note 2)
74	TscH2DIL, TscL2DIL	Hold Time of SDI Data Input to SCK Edge		100	—	ns	
75	TDOR	SDO Data Output Rise Time		—	25	ns	
76	TDOF	SDO Data Output Fall Time		—	25	ns	
77	TssH2doZ	SS ↑ to SDO Output High-Impedand	ce	10	50	ns	
78	TscR	SCK Output Rise Time (Master mod	le)	—	25	ns	
79	TscF	SCK Output Fall Time (Master mode	e)	_	25	ns	
80	TSCH2DOV, TSCL2DOV	SDO Data Output Valid after SCK Edge		—	50	ns	
82	TssL2doV	SDO Data Output Valid after $\overline{\text{SS}}\downarrow$ Edge		_	50	ns	
83	TscH2ssH, TscL2ssH	SS ↑ after SCK Edge		1.5 Tcy + 40	—	ns	

Note 1: Requires the use of Parameter #73A.



Param. No.	Symbol	Characte	ristic	Min	Max	Units	Conditions	
90	TSU:STA	Start Condition	100 kHz mode	4700		ns	Only relevant for Repeated	
		Setup Time	400 kHz mode	600	—		Start condition	
91	THD:STA	Start Condition	100 kHz mode	4000	_	ns	After this period, the first	
		Hold Time	400 kHz mode	600	_		clock pulse is generated	
92	TSU:STO	Stop Condition	100 kHz mode	4700	_	ns		
		Setup Time	400 kHz mode	600	_			
93	THD:STO	Stop Condition	100 kHz mode	4000	_	ns		
		Hold Time	400 kHz mode	600				

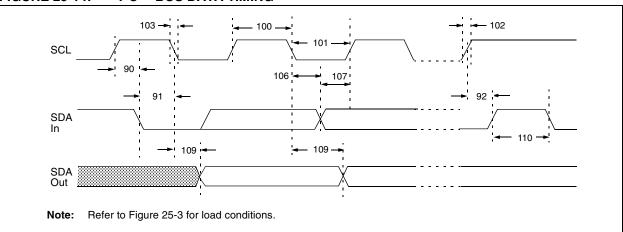


FIGURE 25-14: I²C[™] BUS DATA TIMING

Param. No.	Symbol	Characteris	tic	Min	Max	Units	Conditions
100	Тнідн	Clock High Time	100 kHz mode	4.0		μs	
			400 kHz mode	0.6	_	μs	
			MSSP Module	1.5 TCY	—		
101	TLOW	Clock Low Time	100 kHz mode	4.7	—	μs	
			400 kHz mode	1.3	—	μs	
			MSSP Module	1.5 TCY	—		
102	TR	SDA and SCL Rise Time	100 kHz mode	—	1000	ns	
			400 kHz mode	20 + 0.1 CB	300	ns	CB is specified to be from 10 to 400 pF
103	TF	SDA and SCL Fall Time	100 kHz mode	—	300	ns	
			400 kHz mode	20 + 0.1 CB	300	ns	CB is specified to be from 10 to 400 pF
90	TSU:STA	Start Condition Setup Time	100 kHz mode	4.7		μs	Only relevant for Repeated
			400 kHz mode	0.6		μs	Start condition
91	THD:STA	Start Condition Hold Time	100 kHz mode	4.0	_	μs	After this period, the first clock
			400 kHz mode	0.6	_	μs	pulse is generated
106	THD:DAT	Data Input Hold Time	100 kHz mode	0	—	ns	
			400 kHz mode	0	0.9	μs	
107	TSU:DAT	Data Input Setup Time	100 kHz mode	250	—	ns	(Note 2)
			400 kHz mode	100	—	ns	
92	Tsu:sto	Stop Condition Setup Time	100 kHz mode	4.7	—	μs	
			400 kHz mode	0.6	—	μs	
109	ΤΑΑ	Output Valid from Clock	100 kHz mode	—	3500	ns	(Note 1)
			400 kHz mode		—	ns	
110	TBUF	Bus Free Time	100 kHz mode	4.7	_	μs	Time the bus must be free before
			400 kHz mode	1.3	—	μs	a new transmission can start
D102	Св	Bus Capacitive Loading			400	pF	

TABLE 25-19:	I ² C [™] BUS DATA REQUIREMENTS (SLAVE MODE)
--------------	--

Note 1: As a transmitter, the device must provide this internal minimum delay time to bridge the undefined region (min. 300 ns) of the falling edge of SCL to avoid unintended generation of Start or Stop conditions.

2: A Fast mode I²C[™] bus device can be used in a Standard mode I²C bus system, but the requirement, TSU:DAT ≥ 250 ns, must then be met. This will automatically be the case if the device does not stretch the LOW period of the SCL signal. If such a device does stretch the LOW period of the SCL signal, it must output the next data bit to the SDA line, TR max. + TSU:DAT = 1000 + 250 = 1250 ns (according to the Standard mode I²C bus specification), before the SCL line is released.

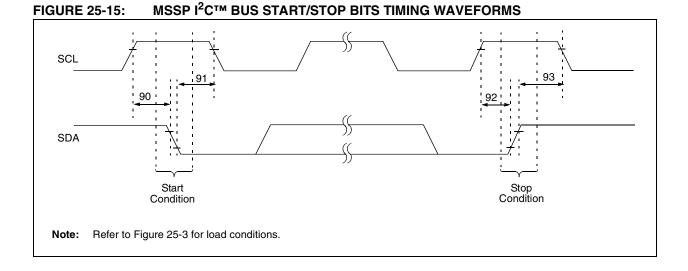
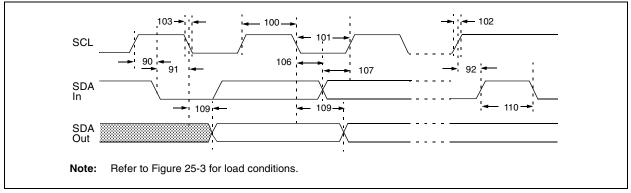


TABLE 25-20:	MSSP I ² C [™] BUS START/STOP BITS REQUIREMENTS
--------------	---

Param. No.	Symbol	Characteristic		Min	Max	Units	Conditions
90	TSU:STA	Start Condition	100 kHz mode	2(Tosc)(BRG + 1)		ns	Only relevant for
		Setup Time	400 kHz mode	2(Tosc)(BRG + 1)	_		Repeated Start
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)	_		condition
91	THD:STA	Start Condition	100 kHz mode	2(Tosc)(BRG + 1)	_	ns	After this period, the
		Hold Time	400 kHz mode	2(Tosc)(BRG + 1)	_		first clock pulse is
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)	_		generated
92	Tsu:sto	Stop Condition	100 kHz mode	2(Tosc)(BRG + 1)	_	ns	
		Setup Time	400 kHz mode	2(Tosc)(BRG + 1)	_		
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)			
93	THD:STO	Stop Condition	100 kHz mode	2(Tosc)(BRG + 1)	_	ns	
		Hold Time	400 kHz mode	2(Tosc)(BRG + 1)	_	1	
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)		1	

Note 1: Maximum pin capacitance = 10 pF for all I^2C^{TM} pins.

FIGURE 25-16: MSSP I²C[™] BUS DATA TIMING



Param. No.	Symbol	Charac	teristic	Min	Max	Units	Conditions
100	Тнідн	Clock High	100 kHz mode	2(Tosc)(BRG + 1)	_	ms	
		Time	400 kHz mode	2(Tosc)(BRG + 1)		ms	
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)		ms	
101	TLOW	Clock Low Time	100 kHz mode	2(Tosc)(BRG + 1)	—	ms	
			400 kHz mode	2(Tosc)(BRG + 1)	_	ms	
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)	_	ms	
102	TR	SDA and SCL	100 kHz mode	_	1000	ns	CB is specified to be from
		Rise Time	400 kHz mode	20 + 0.1 Св	300	ns	10 to 400 pF
			1 MHz mode ⁽¹⁾	_	300	ns	
103	TF	SDA and SCL	100 kHz mode	_	300	ns	CB is specified to be from
		Fall Time	400 kHz mode	20 + 0.1 Св	300	ns	10 to 400 pF
			1 MHz mode ⁽¹⁾	_	100	ns	
90	TSU:STA	Start Condition	100 kHz mode	2(Tosc)(BRG + 1)	—	ms	Only relevant for Repeated
		Setup Time	400 kHz mode	2(Tosc)(BRG + 1)	_	ms	Start condition
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)	_	ms	
91	THD:STA	A Start Condition Hold Time	100 kHz mode	2(Tosc)(BRG + 1)	_	ms	After this period, the first
			400 kHz mode	2(Tosc)(BRG + 1)	_	ms	clock pulse is generated
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)		ms	
106	THD:DAT	Data Input	100 kHz mode	0	_	ns	
		Hold Time	400 kHz mode	0	0.9	ms	
			1 MHz mode ⁽¹⁾	TBD	_	ns	
107	TSU:DAT	Data Input	100 kHz mode	250	_	ns	(Note 2)
		Setup Time	400 kHz mode	100	_	ns	
			1 MHz mode ⁽¹⁾	TBD		ns	
92	Tsu:sto	Stop Condition	100 kHz mode	2(Tosc)(BRG + 1)	—	ms	
		Setup Time	400 kHz mode	2(Tosc)(BRG + 1)	_	ms	
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)	_	ms	
109	ΤΑΑ	Output Valid	100 kHz mode	—	3500	ns	
		from Clock	400 kHz mode	—	1000	ns	
			1 MHz mode ⁽¹⁾	—	—	ns	
110	TBUF	Bus Free Time	100 kHz mode	4.7	—	ms	Time the bus must be free
			400 kHz mode	1.3	_	ms	before a new transmission
			1 MHz mode ⁽¹⁾	TBD		ms	can start
D102	Св	Bus Capacitive L	oading	—	400	pF	

TABLE 25-21: MSSP I²C™ BUS DATA REQUIREMENTS

Legend: TBD = To Be Determined

Note 1: Maximum pin capacitance = 10 pF for all I^2C^{TM} pins.

2: A Fast mode I²C bus device can be used in a Standard mode I²C bus system, but parameter #107 ≥ 250 ns must then be met. This will automatically be the case if the device does not stretch the LOW period of the SCL signal. If such a device does stretch the LOW period of the SCL signal, it must output the next data bit to the SDA line, parameter #102 + parameter #107 = 1000 + 250 = 1250 ns (for 100 kHz mode), before the SCL line is released.

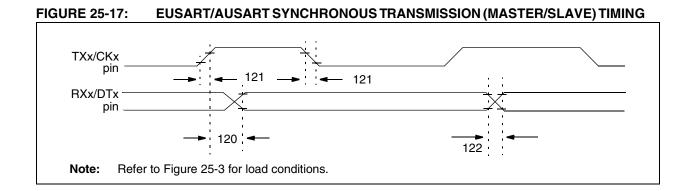


TABLE 25-22: EUSART/AUSART SYNCHRONOUS TRANSMISSION REQUIREMENTS

Param No.	Symbol	Characteristic	Min	Max	Units	Conditions
120	TCKH2DTV	SYNC XMIT (MASTER and SLAVE) Clock High to Data Out Valid	_	40	ns	
121	TCKRF	Clock Out Rise Time and Fall Time (Master mode)	_	20	ns	
122	TDTRF	Data Out Rise Time and Fall Time		20	ns	

FIGURE 25-18: EUSART/AUSART SYNCHRONOUS RECEIVE (MASTER/SLAVE) TIMING

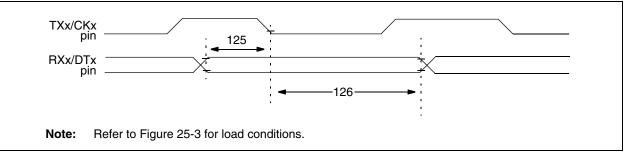


TABLE 25-23: EUSART/AUSART SYNCHRONOUS RECEIVE REQUIREMENTS

Param. No.	Symbol	Characteristic		Max	Units	Conditions
125	TDTV2CKL	<u>SYNC RCV (MASTER and SLAVE)</u> Data Hold before CKx \downarrow (DTx hold time)	10	_	ns	
126	TCKL2DTL	Data Hold after CKx \downarrow (DTx hold time)	15		ns	

Param No.	Symbol	Characteristic	Min	Тур	Мах	Units	Conditions
A01	NR	Resolution	—	_	10	bits	
A03	EIL	Integral Linearity Error	—	_	<±1	LSb	$\Delta VREF \ge 3.0V$
A04	Edl	Differential Linearity Error	—	_	<±1	LSb	$\Delta VREF \ge 3.0V$
A06	EOFF	Offset Error	—	_	<±3	LSb	$\Delta VREF \ge 3.0V$
A07	Egn	Gain Error	_	_	<±3	LSb	$\Delta VREF \ge 3.0V$
A10	—	Monotonicity	Gi	Guaranteed ⁽¹⁾		_	$VSS \le VAIN \le VREF$
A20	$\Delta VREF$	Reference Voltage Range (VREFH – VREFL)	2.0 3	_		V V	VDD < 3.0V $VDD \ge 3.0V$
A21	VREFH	Reference Voltage High	Vss	_	VREFH	V	
A22	VREFL	Reference Voltage Low	Vss – 0.3V		Vdd - 3.0V	V	
A25	VAIN	Analog Input Voltage	VREFL		VREFH	V	
A30	ZAIN	Recommended Impedance of Analog Voltage Source	—		2.5	kΩ	
A50	IREF	VREF Input Current ⁽²⁾		_	5 150	μΑ μΑ	During VAIN acquisition. During A/D conversion cycle.

TABLE 25-24: A/D CONVERTER CHARACTERISTICS: PIC18F85J90 FAMILY (INDUSTRIAL)

Note 1: The A/D conversion result never decreases with an increase in the input voltage and has no missing codes.

2: VREFH current is from RA3/AN3/VREF+ pin or VDD, whichever is selected as the VREFH source. VREFL current is from RA2/AN2/VREF- pin or VSS, whichever is selected as the VREFL source.

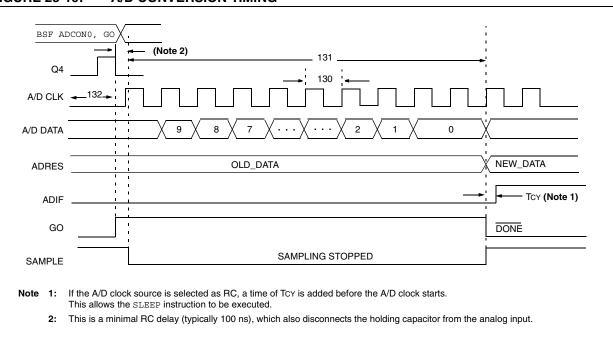


FIGURE 25-19: A/D CONVERSION TIMING

TABLE 25-25: A/D CONVERSION REQUIREMENTS

Param No.	Symbol	Characteristic	Min	Мах	Units	Conditions
130	Tad	A/D Clock Period	0.7	25.0 ⁽¹⁾	μs	Tosc based, VREF \geq 3.0V
			TBD	1	μs	A/D RC mode
131	TCNV	Conversion Time (not including acquisition time) ⁽²⁾	11	12	Tad	
132	TACQ	Acquisition Time ⁽³⁾	1.4	—	μs	-40°C to +85°C
135	Tswc	Switching Time from Convert \rightarrow Sample		(Note 4)		
TBD	TDIS	Discharge Time	0.2	—	μs	

Legend: TBD = To Be Determined

Note 1: The time of the A/D clock period is dependent on the device frequency and the TAD clock divider.

2: ADRES registers may be read on the following TCY cycle.

3: The time for the holding capacitor to acquire the "New" input voltage when the voltage changes full scale after the conversion (VDD to Vss or Vss to VDD). The source impedance (*Rs*) on the input channels is 50Ω.

4: On the following cycle of the device clock.

NOTES:

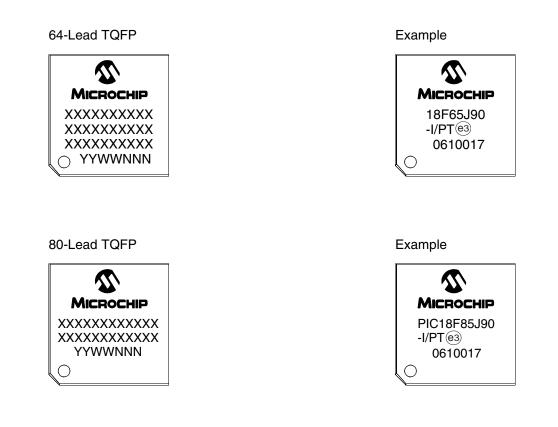
26.0 DC AND AC CHARACTERISTICS GRAPHS AND TABLES

Graphs and tables are not available at this time.

NOTES:

27.0 PACKAGING INFORMATION

27.1 Package Marking Information



Legend:	XXX Y YY WW NNN @3 *	Customer-specific information Year code (last digit of calendar year) Year code (last 2 digits of calendar year) Week code (week of January 1 is week '01') Alphanumeric traceability code Pb-free JEDEC designator for Matte Tin (Sn) This package is Pb-free. The Pb-free JEDEC designator (e3) can be found on the outer packaging for this package.
	be carrie	nt the full Microchip part number cannot be marked on one line, it will d over to the next line, thus limiting the number of available s for customer-specific information.

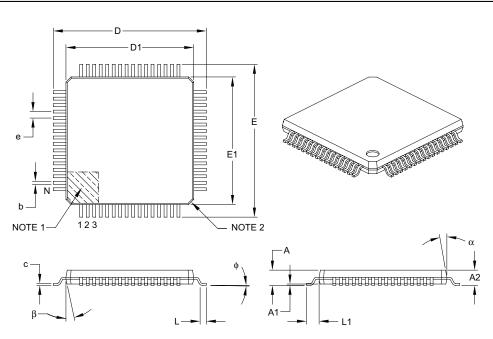
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27.2 Package Details

The following sections give the technical details of the packages.

64-Lead Plastic Thin Quad Flatpack (PT) – 10x10x1 mm Body, 2.00 mm Footprint [TQFP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units	MILLIMETERS			
	MIN	NOM	MAX		
Number of Leads	N		64		
Lead Pitch	е		0.50 BSC		
Overall Height	А	-	-	1.20	
Molded Package Thickness	A2	0.95	1.00	1.05	
Standoff	A1	0.05	-	0.15	
Foot Length	L	0.45	0.60	0.75	
Footprint	L1	1.00 REF			
Foot Angle	¢	0°	3.5°	7°	
Overall Width	E		12.00 BSC		
Overall Length	D		12.00 BSC		
Molded Package Width	E1		10.00 BSC		
Molded Package Length	D1		10.00 BSC		
Lead Thickness	С	0.09	_	0.20	
Lead Width	b	0.17	0.22	0.27	
Mold Draft Angle Top	α	11° 12° 13°			
Mold Draft Angle Bottom	β	11°	12°	13°	

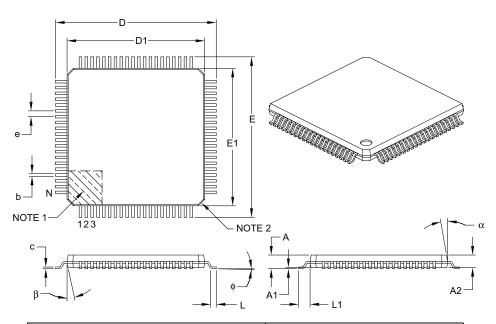
Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. Chamfers at corners are optional; size may vary.
- 3. Dimensions D1 and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.25 mm per side.
- 4. Dimensioning and tolerancing per ASME Y14.5M.
 - BSC: Basic Dimension. Theoretically exact value shown without tolerances.
 - REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-085B

80-Lead Plastic Thin Quad Flatpack (PT) – 12x12x1 mm Body, 2.00 mm Footprint [TQFP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units	MILLIMETERS		
	Dimension Limits	MIN	NOM	MAX
Number of Leads	N	80		
Lead Pitch	е	0.50 BSC		
Overall Height	А	-	-	1.20
Molded Package Thickness	A2	0.95	1.00	1.05
Standoff	A1	0.05	-	0.15
Foot Length	L	0.45	0.60	0.75
Footprint	L1	1.00 REF		
Foot Angle	φ	0°	3.5°	7°
Overall Width	E	14.00 BSC		
Overall Length	D	14.00 BSC		
Molded Package Width	E1	12.00 BSC		
Molded Package Length	D1	12.00 BSC		
Lead Thickness	С	0.09	_	0.20
Lead Width	b	0.17	0.22	0.27
Mold Draft Angle Top	α	11°	12°	13°
Mold Draft Angle Bottom	β	11°	12°	13°

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

2. Chamfers at corners are optional; size may vary.

3. Dimensions D1 and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.25 mm per side.

- 4. Dimensioning and tolerancing per ASME Y14.5M.
 - BSC: Basic Dimension. Theoretically exact value shown without tolerances. REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-092B

NOTES:

APPENDIX A: REVISION HISTORY

Revision A (July 2006)

Original data sheet for PIC18F85J90 family devices.

Revision B (March 2007)

Updated power-down and supply-current electrical characteristics and package details illustrations.

APPENDIX B: MIGRATION BETWEEN HIGH-END DEVICE FAMILIES

Devices in the PIC18F85J90 and PIC18F8490 families are very similar in their functions and feature sets. However, there are some potentially important differences which should be considered when migrating an application across device families to achieve a new design goal. These are summarized in Table B-1. The areas of difference, which could be a major impact on migration, are discussed in greater detail later in this section.

TABLE B-1: NOTABLE DIFFERENCES BETWEEN PIC18F8490 AND PIC18F85J90 FAMILIES

Characteristic	PIC18F85J90 Family	PIC18F8490 Family	
Operating Frequency	40 MHz @ 2.15V	40 MHz @ 4.2V	
Supply Voltage	2.0V-3.6V, Dual Voltage Requirement	2.0V-5.5V	
Operating Current	Low	Lower	
Program Memory Size (maximum)	32 Kbytes	16 Kbytes	
Program Memory Endurance	10,000 Write/Erase Cycles (typical)	100,000 Write/Erase Cycles (typical)	
Program Memory Retention	20 Years (minimum)	40 Years (minimum)	
Programming Time (Normalized)	43.8 µs/byte (2.8 ms/64-byte block)	15.6 μs/byte (1 ms/64-byte block)	
I/O Sink/Source at 25 mA	PORTB and PORTC Only	All Ports	
Input Voltage Tolerance on I/O Pins	5.5V on Digital Only Pins	VDD on All I/O Pins	
I/O	67	66	
LCD Outputs (maximum pixels, segments x commons)	192	192	
LCD Bias Generation	4 Modes	1 Mode	
LCD Voltage Regulator	Implemented; Includes Voltage Boost	Not Available	
Pull-ups	PORTB, PORTD, PORTE and PORTJ	PORTB	
Open-Drain Output Option	Available on USARTs, SPI and CCP Output Pins	Not Available	
Oscillator Options	Limited Primary Options (EC, HS, PLL); Flexible Internal Oscillator (INTOSC and INTRC)	More Primary Options (EC, HS, XT, LP, RC, PLL); Flexible Internal Oscillator (INTOSC and INTRC)	
Programming Entry	Low Voltage, Key Sequence	VPP and LVP	
Code Protection	Single Block, All or Nothing	Multiple Code Protection Blocks	
Configuration Words	Stored in Last 4 Words of Program Memory space	Stored in Configuration Space, Starting at 300000h	
Start-up Time from Sleep	200 μs (typical) 10 μs (typical) with Voltage Regulator Disabled	10 μs (typical)	
Power-up Timer	Always on	Configurable	
Data EEPROM	Use Self-Programming	Not Available	
BOR	Simple BOR with Voltage Regulator	Separate Programmable BOR	
LVD	Integrated with Voltage Regulator	Separate Programmable Module	
A/D Channels		12	
A/D Calibration	Self-Calibration Feature	Software Look-up Table	
In-Circuit Emulation	Not available Software Look-up Table		

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B.1 Power Requirement Differences

The most significant difference between the PIC18F85J90 and PIC18F8490 device families is the power requirements. PIC18F85J90 family devices are designed on a smaller process. This results in lower maximum voltage and higher leakage current.

The operating voltage range for PIC18F85J90 devices is 2.0V to 3.6V. In addition, these devices have split power requirements: one for the core logic and one for the I/O. One of the VDD pins is separated for the core logic supply (VDDCORE). This pin has specific voltage and capacitor requirements as described in **Section 25.0 "Electrical Characteristics"**.

B.2 Oscillator Differences

PIC18F8490 and PIC18F85J90 family devices share a similar range of oscillator options. The major difference is that PIC18F85J90 family devices support a smaller number of primary (external) oscillator options, namely HS and EC Oscillator modes.

While both device families have an internal PLL that can be used with the primary oscillators, the PLL for the PIC18F85J90 family is not enabled as a device configuration option. Instead, it must be enabled in software.

The clocking differences should be considered when making a conversion between the PIC18F8490 and PIC18F85J90 device families.

B.3 LCD Module

When converting an LCD application between the PIC18F85J90 and the PIC18F8490 families, the following things must be considered:

- Available Segments: The module for PIC18F65J90 devices supports 33 segments, as opposed to 32 segments in PIC18F6490 devices. (The 80-pin devices of both families support 48 segments. All devices support 4 commons.)
- **Bias Generation:** The PIC18F85J90 version of the module also incorporates its own independent voltage regulator, which supports 4 circuit configurations for bias generation, voltage boost to support displays that operate above device VDD and software contrast control.

- Additional LCD Function Pins: The PIC18F85J90 family of devices adds 3 additional LCD function pins in comparison to the PIC18F8490 family. The additional pins are associated with LCD bias generation:
 - LCDBIAS0 (RG0)
 - VLCAP1 (RG2)
 - VLCPA2 (RG3)
- Segment Assignments: Eight of the LCD segment functions have been relocated to different I/O pins than in PIC18F8490 devices. These segments are listed in Table B-2.
- Other Considerations: In all LCD applications, the connections of PIC18F85J90 devices to external components for LCD bias generation are different than PIC18F8490 devices. The addition of the LCDBIAS0 output requires that this pin be included in bias component configurations. A more complete discussion is provided in Section 15.3 "LCD Bias Generation". The simultaneous use of the external Timer1 oscillator and Segment 32 is not allowed in PIC18F85J90 devices, since these functions are shared on the same pin.

TABLE B-2:	ASSIGNMENTS OF MOVED
	LCD SEGMENTS

RA2	RC4
RA3	RC3
RF0	RA1
RG3	RC6
RG2	RC7
RG0	RB5
RG0	RB0
RJ0	RC1
	RA3 RF0 RG3 RG2 RG0 RG0

Note: Refer to the pinout diagrams for pin locations of I/O ports.

B.4 Pin Differences

Besides the LCD pinout differences already described, there are other differences in the pinouts between the PIC18F85J90 and the PIC18F8490 families:

- Input voltage tolerance
- · Output current capabilities
- Available I/O

Pins on the PIC18F85J90 that have digital only input capability will tolerate voltages up to 5.5V, and are thus, tolerant to voltages above VDD. Table 9-1 in **Section 9.1 "I/O Port Pin Capabilities**" contains the complete list.

In addition to input differences, there are output differences as well. PIC18F85J90 devices have three classes of pin output current capability: high, medium and low. Not all I/O pins can source or sink equal levels of current. Only PORTB and PORTC support the 25 mA source/sink capability that is supported by all output pins on the PIC18F8490. Table 9-1 in **Section 9.1 "I/O Port Pin Capabilities**" contains the complete list of output capabilities.

Finally, the pins associated with the CCP, EUSART/AUSART and SPI peripherals can be configured by the user as open-drain outputs. This allows for simpler interfacing with external devices operating at higher voltages. This capability is not directly equivalent to any feature on the PIC18F8490 family.

There are also differences in the implementation of some ports on PIC18F85J90 devices. While the total number of general purpose I/O pins are very similar (67 vs. 66), the implementation of individual pins has notable differences:

- The MCLR pin is dedicated only to MCLR and cannot be configured as an input (RG5) as it can on PIC18F8490 devices.
- RF0 does not exist on PIC18F85J90 devices.
- RE0, RE1 and RE3 are implemented on PIC18F85J90 devices, but not PIC18F8490 devices.

All of these pin differences (including power pin differences) should be accounted for when making a conversion between PIC18F8490 and PIC18F85J90 devices.

B.5 Other Peripherals

Peripherals must also be considered when making a conversion between the PIC18F85J90 and the PIC18F8490 families:

- **A/D Converter:** The converter for PIC18F85J90 devices require a calibration step prior to normal operation.
- Data EEPROM: PIC18F85J90 devices do not have this module but offer self-programming capability.
- **BOR:** PIC18F85J90 devices do not have a programmable BOR. Simple brown-out capability is provided through the use of the internal voltage regulator.
- LVD: PIC18F85J90 devices do not have this module. A limited, fixed setpoint capability is provided through the use of the internal voltage regulator.

NOTES:

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Device	PIC18F63J90/64J90/65J90 ⁽¹⁾ , PIC18F83J90/84J90/85J90 ⁽¹⁾ , PIC18F63J90/64J90/65J90T ⁽²⁾ , PIC18F83J90/84J90/85J90T ⁽²⁾	
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