

PIC16(L)F1777/8/9

28/40/44-Pin, 8-Bit Flash Microcontroller

Description

PIC16(L)F1777/8/9 microcontrollers feature a high level of integration of intelligent analog and digital peripherals for a wide range of applications, such as lighting, power supplies, battery charging, motor control and other general purpose applications. These devices deliver multiple op amps, 5-/10-bit DACs, high-speed comparators, 10-bit ADC, 10-/16-bit PWMs, programmable ramp generator (PRG) and other peripherals that can be connected internally to create closed-loop systems without using pins or the printed circuit board (PCB) area. The 10-/16-bit PWMs, digital signal modulators and tri-state output op amp can be used together to create a LED dimming engine for lighting applications. The peripheral pin select (PPS) functionality provides flexibility, eases PCB layout and peripheral utilization by allowing digital peripheral pin mapping to an I/O.

Core Features

- C Compiler Optimized RISC Architecture
- Only 49 Instructions
- Operating Speed:
- DC 32 MHz clock input
- 125 ns minimum instruction cycle
- Interrupt Capability
- 16-Level Deep Hardware Stack
- Five 8-Bit Timers
- Three 16-Bit Timers
- Low-Current Power-on Reset (POR)
- Configurable Power-up Timer (PWRT)
- · Brown-out Reset (BOR) with Selectable Trip Point
- Extended Watchdog Timer (EWDT):
 - Low-power 31 kHz WDT
 - Software selectable prescaler
 - Software selectable enable

Memory

- Up to 28 Kbytes Program Flash Memory (PFM)
- · Up to 2 Kbytes Data RAM
- · Direct, Indirect and Relative Addressing modes
- High-Endurance Flash (HEF):
 - 128B of nonvolatile data storage
 - 100K Erase/Write cycles

Operating Characteristics

- Operating Voltage Range:
 - 1.8V to 3.6V (PIC16LF1777/8/9)
 - 2.3V to 5.5V (PIC16F1777/8/9)
- Temperature Range:
 - Industrial: -40°C to 85°C
 - Extended: -40°C to 125°C

eXtreme Low-Power (XLP) Features

- Sleep mode: 50 nA @ 1.8V, typical
- Watchdog Timer: 500 nA @ 1.8V, typical
- Secondary Oscillator: 500 nA @ 32 kHz
- Operating Current:
 - 8 uA @ 31 kHz, 1.8V, typical
 - 32 uA/MHz @ 1.8V, typical

Intelligent Analog Peripherals

- 10-Bit Analog-to-Digital Converter (ADC):
 - Up to 28 external channels
 - Conversion available during Sleep
- Four Operational Amplifiers (OPA):
 - Selectable internal and external channels
 Tri-state output
 - Part of LED dimming engine
- Selectable internal and external channels
- Eight High-Speed Comparators (HS Comp):
 - Up to nine external inverting inputs
 - Up to 12 external non-inverting inputs
 - Fixed Voltage Reference at inverting and non-inverting input(s)
 - Comparator outputs externally accessible
- Digital-to-Analog Converters (DAC):
- Four 10-bit resolution DACs
- 10-bit resolution, rail-to-rail
- Conversion during Sleep
- Internal connections to ADCs and HS Comparators
- · Voltage Reference:
 - Fixed Voltage Reference (FVR)
 - 1.024V, 2.048V and 4.096V output levels
- Zero-Cross Detector (ZCD):
 - Detect high-voltage AC signal
- Four Programmable Ramp Generators (PRG):
 - Slope compensation
 - Ramp generation
- High-Current Drive I/Os:
 - Up to 100 mA sink or source @ 5V

Digital Peripherals

- Four Configurable Logic Cells (CLC):
 Integrated combinational and state logic
- Four Complementary Output Generators (COG):
 Push-pull, Full-Bridge and Steering modes
- Fust-pull, Pull-Bridge and Steering modes
 Four Capture/Compare/PWM (CCP) Modules
- Pulse-Width Modulator (PWM):
- Four 16-bit PWMs
- Independent timers
- Multiple output modes (Edge-, Center-Aligned, set and toggle on register match)
- User settings for phase, duty cycle, period, offset and polarity
- 16-bit timer capability
- Three 10-bit PWMs
- Digital Signal Modulator (DSM):
 - Modulates a carrier signal with a digital data to create custom carrier synchronized output waveforms
 - Part of LED dimming engine
- Peripheral Pin Select (PPS):

- I/O remapping of digital peripherals
- Serial Communications:
 - Enhanced USART (EUSART)
 - SPI, I²C, RS-232, RS-485, LIN compatible
 - Auto-Baud Detect, auto-wake-up on start
- Up to 25 I/O Pins:
 - Individually programmable pull-ups
 - Slew rate control
 - Interrupt-on-change with edge-select

Clocking Structure

- Precision Internal Oscillator:
 - ±1% at calibration
 - Selectable frequency range 32 MHz to 31 kHz
- 31 kHz Low-Power Internal Oscillator
- 4x Phase-Locked Loop (PLL) for up to 32 MHz Internal Operation
- External Oscillator Block with Three External Clock modes up to 32 MHz

TABLE 1:	PIC	16(L)	F1//3	5/6/7/	8/9 F		ILY	IYP	ES													
Device	Data Sheet Index	Program Flash Memory (bytes)	Program Flash Memory (word)	High Endurance Flash (B)	Data SRAM (Bytes)	I/O Pins ⁽¹⁾	8-Bit/16-Bit Timers	High-Speed Comparator	10-bit ADC (ch)	5/10-bit DAC	ССР	10-bit/16-bit PWM	900	CLC	Op Amp	Zero Cross Detect	Programmable Ramp Gen	High-Current I/Os	Peripheral Pin Select	EUSART	I ² C/SPI	Debug ⁽²⁾
PIC16(L)F1773	(A)	7K	4K	128	512	25	5/3	6	17	3/3	3	3/3	3	4	3	1	3	2	Y	1	1	Ι
PIC16(L)F1776	(A)	14K	8K	128	1K	25	5/3	6	17	3/3	3	3/3	3	4	3	1	3	2	Υ	1	1	Ι
PIC16(L)F1777	(B)	14K	8K	128	1K	36	5/3	8	28	4/4	4	4/4	4	4	4	1	4	2	Υ	1	1	Ι
PIC16(L)F1778	(B)	28K	16K	128	2K	25	5/3	6	17	3/3	3	3/3	3	4	3	1	3	2	Υ	1	1	Ι
PIC16(L)F1779	(B)	28K	16K	128	2K	36	5/3	8	28	4/4	4	4/4	4	4	4	1	4	2	Υ	1	1	Ι

Note 1: One pin is input-only.

2: I – Debugging integrated on chip.

Data Sheet Index:

A: DS40001810

0001810 PIC16(L)F1773/6 Data Sheet, 28-Pin, 8-bit Flash Microcontrollers

B: DS40001819

PIC16(L)F1777/8/9 Data Sheet, 28/40/44-Pin, 8-bit Flash Microcontrollers

Note: For other small form-factor package availability and marking information, please visit http://www.microchip.com/packaging or contact your local sales office.

 TABLE 1:
 PIC16(L)F1773/6/7/8/9 FAMILY TYPES

TABLE 2: PACKAGES

Packages	SPDIP	PDIP	SOIC	SSOP	UQFN	QFN	TQFP
PIC16(L)F1778	•		٠	٠	٠		
PIC16(L)F1777/9		•			٠	٠	•

Note: Pin details are subject to change.

PIC16(L)F1777/8/9

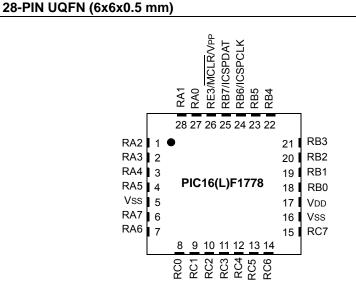
28-PIN SPDIP, SOIC, SSOP

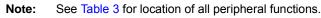
PIN DIAGRAMS

FIGURE 1:

FIGURE 2:

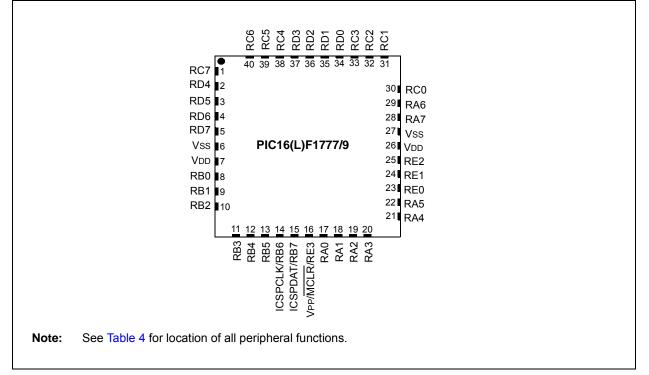
	VPP/MCLR/RE3	\bigcirc	28	RB7/ICSPDAT
	RA0 2		27	RB6/ICSPCLK
	RA1 🗌 3		26	RB5
	RA2 4		25	RB4
	RA3 🛛 5		24	RB3
	RA4 🗌 6	8	23	RB2
	RA5 🗌 7	PIC16(L)F1778	22	RB1
	Vss 🗌 8	Ľ	21	RB0
	RA7 🗌 9	16(20	VDD
	RA6 🗌 10	Š	19	Vss
	RC0 🗌 11	ш.	18	RC7
	RC1 🗌 12		17	RC6
	RC2 13		16] RC5
	RC3 🗌 14		15	RC4
Note: See Table 3 for loc	cation of all peripheral fun	ctions.		



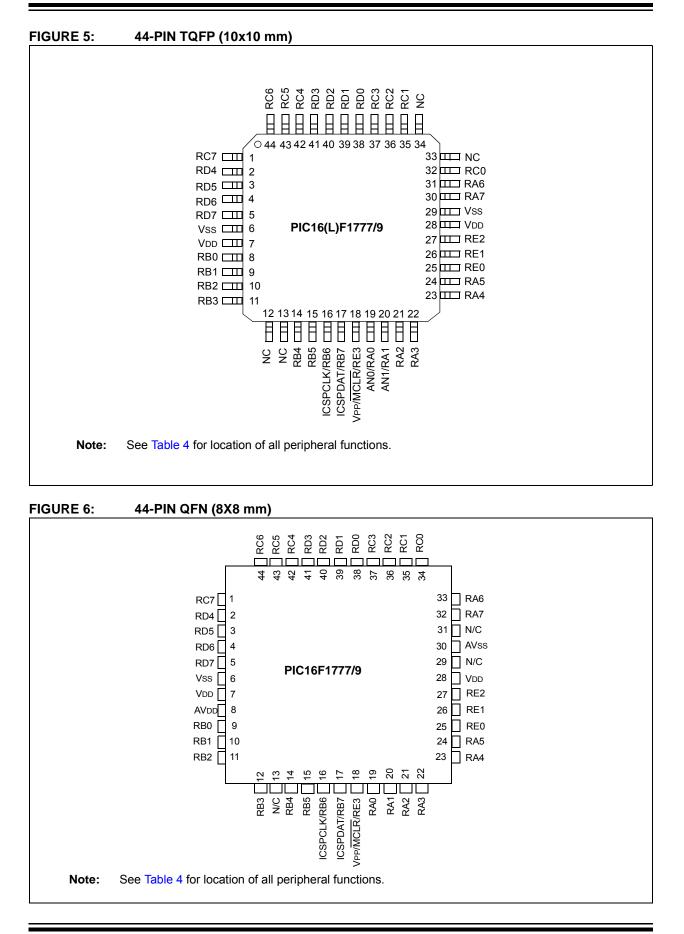


SURE 3:	40-PIN PDIP			
	VPP/MCLR/RE3	\bigcirc	40 RB7/ICSPDAT	
	RA0 2		39 RB6/ICSPCLK	
			38 RB5	
	RA1 [] 3			
	RA2 [] 4		37 RB4	
	RA3 5		36 RB3	
	RA4 6		35 RB2	
	RA5 🗌 7		34 RB1	
	RE0 8	-	33 RB0	
	RE1 🗌 9	6/1		
	RE2 🗌 10	11	31 Vss	
	VDD 🗌 11	Ľ,	30 RD7	
	VSS [] 12	1)9	29 RD6	
	RA7 🗌 13	PIC16(L)F177/9	28 RD5	
	RA6 🗌 ₁₄	<u>а</u>	27 RD4	
	RC0 15		26 RC7	
	RC1 16		25 RC6	
	RC2 17		24 RC5	
	RC3 18		23 RC4	
	RD0 19		22 RD3	
	RD1 20		21 RD2	
-				
Note: See	e Table 4 for location of all peripheral f	unctions.		

FIGURE 4: 40-PIN UQFN (5x5x0.5 mm)



PIC16(L)F1777/8/9



PIN ALLOCATION TABLES

TABLE 3: 28-PIN ALLOCATION TABLE (PIC16(L)F1778)

Q	28-Pin SPDIP/SOIC/SSOP	28-Pin UQFN	ADC	VREF	DAC	Op Amp	Comparator	ZCD	PRG	Timers	MWG	ССР	coc	CLC	Modulator	EUSART	dssw	Interrupt	Pull-ups	High Current	Basic
RA0	2	27	AN0	_	_	_	C1IN0- C2IN0- C3IN0- C4IN0- C5IN0- C6IN0-	_	_	I	_	l	_	CLCIN0 ⁽¹⁾	I			IOC	Y	_	—
RA1	3	28	AN1	_	_	OPA1OUT OPA2IN1+ OPA2IN1-	C1IN1- C2IN1- C3IN1- C4IN1-	_	PRG1IN0 PRG2IN1	_	_	_	_	CLCIN1 ⁽¹⁾	_	_	_	IOC	Y	_	_
RA2	4	1	AN2 Vref-	DAC1REF0- DAC2REF0- DAC3REF0- DAC4REF0- DAC5REF0- DAC7REF0-	DAC1OUT1	_	C1IN0+ C2IN0+ C3IN0+ C4IN0+ C5IN0+ C6IN0+	_	_	_	_	_	_	_	_	_	_	IOC	Y	_	_
RA3	5	2	AN3 VREF+	DAC1REF0+ DAC2REF0+ DAC3REF0+ DAC4REF0+ DAC5REF0+ DAC7REF0+	_	_	C1IN1+	_	_	Ι			_	_	MD1CL ⁽¹⁾		-	IOC	Y	_	_
RA4	6	3	_	_	DAC4OUT1	OPA1IN0+	_	—	PRG1R ⁽¹⁾	T0CKI	-	_	_	_	MD1CH ⁽¹⁾	_	_	IOC	Y	—	_
RA5	7	4	AN4	_	DAC2OUT1	OPA1IN0-	_	—	PRG1F ⁽¹⁾	_	_	_	_	_	MD1MOD ⁽¹⁾	—	SS	IOC	Y	_	—
RA6	10	7	_	-	_	_	C6IN1+	-	—	_	-	_	—	—	—	—	_	IOC	Y	_	OSC2 CLKOUT
RA7	9	6	_	_	_	_	_	-	_	_	_	_	_	_	-	_	_	IOC	Y	-	OSC1 CLKIN
RB0	21	18	AN12	-	_	_	C2IN1+	ZCD	—	_	-	_	COG1IN ⁽¹⁾	-	-	—	_	IOC INT	Y	HIB0	-
RB1	22	19	AN10	-	-	OPA2OUT OPA1IN1+ OPA1IN1-	C1IN3- C2IN3- C3IN3- C4IN3-	_	PRG2IN0 PRG1IN1	_	_	_	COG2IN ⁽¹⁾	_	_	_	_	IOC	Y	HIB1	_
RB2	23	20	AN8	—	DAC3OUT1	OPA2IN0-	_	—	_	—	—	_	COG3IN ⁽¹⁾	_	_	—	_	IOC	Y	—	—
RB3	24	21	AN9	—	—	OPA2IN0+	C1IN2- C2IN2- C3IN2-	_	_	_	_	_	_	_	MD3CL ⁽¹⁾	_	_	IOC	Y	_	-

PIC16(L)F1777/8/9

Note 1: 2: 3:

Default peripheral input. Input can be moved to any other pin with the PPS input selection register. All pin outputs default to PORT latch data. Any pin can be selected as a digital peripheral output with the PPS output selection registers. These peripheral functions are bidirectional. The output pin selections must be the same as the input pin selections.

TABLE 3: 28-PIN ALLOCATION TABLE (PIC16(L)F1778) (CONTINUED)

			201		•••••••		(· · • · •	·-/·		••••••		7									
0/I	28-Pin SPDIP/SOIC/SSOP	28-Pin UQFN	ADC	VREF	DAC	Op Amp	Comparator	ZCD	PRG	Timers	PWM	ССР	COG	CLC	Modulator	EUSART	MSSP	Interrupt	Pull-ups	High Current	Basic
RB4	25	22	AN11	_	-	_	C3IN1+	_	_	T5G	_	-	_	_	MD3CH ⁽¹⁾	_	_	IOC	Y		_
RB5	26	23	AN13	DAC5REF1- DAC7REF1-	—	—	C4IN2-	—	—	T1G	_	CCP7 ⁽¹⁾	-	-	MD3MOD ⁽¹⁾	_	-	IOC	Y	—	—
RB6	27	24	—	DAC5REF1+ DAC7REF1+	-	—	C4IN1+	—	-	—		—	-	CLCIN2 ⁽¹⁾	-		-	IOC	Y	—	ICSPCLK
RB7	28	25	_	_	DAC1OUT2 DAC2OUT2 DAC3OUT2 DAC4OUT2 DAC5OUT2 DAC7OUT2	_	C5IN1+	_	_	T6IN ⁽¹⁾	_	_	_	CLCIN3 ⁽¹⁾	—	_	_	IOC	Y	_	ICSPDAT
RC0	11	8	-	-	DAC5OUT1	-	_	-	-	T1CKI ⁽¹⁾ T3CKI ⁽¹⁾ T3G ⁽¹⁾ SOSCO	_	_		-	—	_	-	IOC	Y	-	-
RC1	12	9	_	_	DAC7OUT1	_	—	—	PRG2R ⁽¹⁾	SOSCI	-	CCP2 ⁽¹⁾	_	_	—		_	IOC	Y	—	—
RC2	13	10	AN14	—	—	—	C5IN2- C6IN2-	_	PRG2F ⁽¹⁾	T5CKI	l	CCP1 ⁽¹⁾		—	-	l	—	IOC	Y	_	—
RC3	14	11	AN15	_	_	_	C1IN4- C2IN4- C3IN4- C4IN4- C5IN4- C6IN4-	_	_	T2IN ⁽¹⁾	_	_	_	_	MD2CL ⁽¹⁾	_	SCL	IOC	Y	_	_
RC4	15	12	AN16	—	-	—	C5IN3- C6IN3-	_	PRG3R ⁽¹⁾	T8IN ⁽¹⁾	l	—		—	MD2CH ⁽¹⁾	I	SDA	IOC	Y	-	—
RC5	16	13	AN17	_	-	OPA3IN0+	_	—	PRG3F ⁽¹⁾	T4IN ⁽¹⁾	_	—	_	_	MD2MOD ⁽¹⁾	_	-	IOC	Y	_	—
RC6	17	14	AN18	_	-	OPA3OUT	C5IN1- C6IN1-	-	PRG3IN0	—		—		-	-		—	IOC	Y	-	—
RC7	18	15	AN19	_	_	OPA3IN0-	—	_	_	_	1	_	-	_	—		_	IOC	Y	_	—
RE3	1	26	—	-	-	—	—	_	—	—		—	_	-	-		—	IOC		—	MCLR VPP
Vdd	20	17	_	—	—	—	_		—	_	Ι	—		—	—		—	—			Vdd
Vss	8	5	_	_	-	—	_	—	—	_	_	_	_	_	—	_	—	_	_	_	Vss
Vss	19	16	_	_	-	—	_	—	—	—	—	—	—	—	—	—	-	—	—	—	Vss

1: 2: 3: Note

Default peripheral input. Input can be moved to any other pin with the PPS input selection register. All pin outputs default to PORT latch data. Any pin can be selected as a digital peripheral output with the PPS output selection registers.

TABLE 3: 28-PIN ALLOCATION TABLE (PIC16(L)F1778) (CONTINUED)

	-						• • •	• •	-/ (-	,									
NO	28-Pin SPDIP/SOIC/SSOP	28-Pin UQFN	ADC	VREF	DAC	Op Amp	Comparator	ZCD	PRG	Timers	MMd	CCP	000	CLC	Modulator	EUSART	dssm	Interrupt	Pull-ups	High Current	Basic
OUT ⁽²⁾	—	—	_	-	_	_	C1OUT C2OUT C3OUT C4OUT C5OUT C6OUT		_	_	PWM3 PWM4 PWM5 PWM6 PWM9 PWM11	CCP1 CCP2 CCP7	COG1A COG1B COG1C COG1D COG2A COG2B COG2C COG2D COG3A COG3B COG3C COG3D	CLC1OUT CLC2OUT CLC3OUT CLC4OUT	MD10UT MD20UT MD30UT	DT ⁽³⁾ ТХ СК	SDO SDA ⁽³⁾ SCK SCL ⁽³⁾	_	_	_	_

Note 1:

Default peripheral input. Input can be moved to any other pin with the PPS input selection register. All pin outputs default to PORT latch data. Any pin can be selected as a digital peripheral output with the PPS output selection registers. These peripheral functions are bidirectional. The output pin selections must be the same as the input pin selections. 2: 3:

DS40	
0018	
19B-p;	
age 1	
0	

TABLE 4:

40-Pin (U)QFN 44-Pin QFN 44-Pin TQFP High Current 40-Pin PDIP Comparator Modulator EUSART Interrupt Pull-ups Amp Timers Basic PWM MSSP ADC VREF DAC СС ZCD PRG СCР 000 õ å RA0 AN0 C1IN0-CLCIN0⁽¹⁾ IOC 2 17 19 19 _ _ _ _ Υ _ _ _ _ _ _ _ _ C2IN0-C3IN0-C4IN0-C5IN0-C6IN0-C7IN0-C8IN0-C1IN1-CLCIN1(1) RA1 3 18 20 20 AN1 OPA1OUT PRG1IN0 IOC Υ _ _ _ _ _ _ _ _ _ _ _ _ OPA2IN1+ C2IN1-PRG2IN1 OPA2IN1-C3IN1-C4IN1-RA2 4 19 21 21 AN2 DAC1REF0-DAC10UT1 C1IN0+ IOC Υ _ _ _ _ _ _ _ _ _ DAC2REF0-C2IN0+ DAC3REF0-C3IN0+ DAC4REF0-C4IN0+ DAC5REF0-C5IN0+ DAC6REF0-C6IN0+ DAC7REF0-C7IN0+ DAC8REF0-C8IN0+ RA3 5 20 22 22 AN3 DAC1REF0+ C1IN1+ MD1CL⁽¹⁾ IOC Υ _ _ _ _ _ _ _ _ _ _ _ _ _ DAC2REF0+ DAC3REF0+ DAC4REF0+ DAC5REF0+ DAC6REF0+ DAC7REF0+ DAC8REF0+ RA4 6 21 23 23 OPA1IN0+ PRG1R⁽¹⁾ MD1CH⁽¹⁾ IOC Υ _ _ _ _ _ _ _ _ _ _ _ _ RA5 7 22 24 24 AN4 DAC2OUT1 OPA1IN0-PRG1F⁽¹⁾ MD1MOD⁽¹⁾ SS IOC Υ _ _ _ _ _ _ _ _ _ _ _ 14 RA6 29 31 33 C6IN1+ _ _ IOC Υ OSC2 _ _ _ _ _ _ _ _ _ _ _ _ _ CLKOUT RA7 13 28 30 32 _ _ _ _ _ _ _ _ _ _ _ IOC Υ _ OSC1 _ _ _ _ CLKIN CCP8⁽¹⁾ 33 ZCD COG1IN⁽¹⁾ MD4CL⁽¹⁾ RB0 8 8 9 AN12 C2IN1+ IOC Υ HIB0 _ _ _ _ _ _ _ _ _ INT COG2IN⁽¹⁾ RB1 34 MD4CH⁽¹⁾ 9 9 10 AN10 OPA2OUT C1IN3-PRG2IN0 IOC Υ HIB1 _ _ _ _ _ _ _ _ _ _ OPA1IN1+ C2IN3-PRG1IN1 OPA1IN1-C3IN3-PRG4R⁽¹⁾ C4IN3-RB2 35 10 DAC3OUT1 OPA2IN0-PRG4F⁽¹⁾ COG3IN⁽¹⁾ MD4MOD⁽¹⁾ 10 11 AN8 _ _ _ _ _ _ _ _ IOC Υ _ 36 RB3 11 11 12 AN9 OPA2IN0+ C1IN2-MD3CL⁽¹⁾ IOC _ _ _ _ _ _ _ _ _ Υ _ _ C2IN2-C3IN2-RB4 37 12 14 14 AN11 _ _ _ C3IN1+ _ _ _ _ MD3CH⁽¹⁾ _ _ IOC Υ _ _ _ _ Note

1: Default peripheral input. Input can be moved to any other pin with the PPS input selection register.

2: All pin outputs default to PORT latch data. Any pin can be selected as a digital peripheral output with the PPS output selection registers.

3: These peripheral functions are bidirectional. The output pin selections must be the same as the input pin selections.

40/44-PIN ALLOCATION TABLE (PIC16(L)F1777/9)

TABI	_E 4	1:	4	0/44	1-PIN	ALLOCA	TION TA	BLE (PI	C16(L)	F17	(1//9) (C		NUED)		•	1				1		
0/1	40-Pin PDIP	40-Pin (U)QFN	44-Pin TQFP	44-Pin QFN	ADC	VREF	DAC	Op Amp	Comparator	CD	PRG	Timers	PWM	ССР	900	CLC	Modulator	EUSART	MSSP	Interrupt	Squ-lluq	High Current	Basic
RB5	38	13	15	15	AN13	DAC5REF1- DAC7REF1-	-	-	C4IN2-	_	-	-	-	CCP7 ⁽¹⁾	_	-	MD3MOD ⁽¹⁾	_	—	IOC	Y	-	—
RB6	39	14	16	16	-	DAC5REF1+ DAC7REF1+	-	-	C4IN1+	_	-	-	_	-	_	CLCIN2 ⁽¹⁾	—	_	—	IOC	Y	_	ICSPCLK
RB7	40	15	17	17	_	_	DAC10UT2 DAC20UT2 DAC30UT2 DAC40UT2 DAC50UT2 DAC60UT2 DAC60UT2 DAC70UT2 DAC80UT2	_	C5IN1+		_	T6IN ⁽¹⁾	_	_	_	CLCIN3 ⁽¹⁾	_	_	_	IOC	Y	_	ICSPDAT
RC0	15	30	32	34	-	_	DAC5OUT1		-	_	-	T1CKI ⁽¹⁾ T3CKI ⁽¹⁾ T3G ⁽¹⁾ SOSCO	-		_	-	_	_	-	IOC	Y	-	-
RC1	16	31	35	35	_	—	DAC7OUT1	_	_	-	PRG2R ⁽¹⁾	SOSCI	_	CCP2 ⁽¹⁾	_	_	_	_	_	IOC	Y	_	_
RC2	17	32	36	36	AN14	—	-	—	C5IN2- C6IN2-	-	PRG2F ⁽¹⁾	-	-	CCP1 ⁽¹⁾	-	—	-		—	IOC	Y	—	—
RC3	18	33	37	37	AN15	_	_	_			_	T2IN ⁽¹⁾	_	_		_	MD2CL ⁽¹⁾		SCL	IOC	Y	_	_
RC4	23	38	42	42	AN16	—	-	—	C5IN3- C6IN3-	-	PRG3R ⁽¹⁾	T8IN ⁽¹⁾	-	-	_	—	MD2CH ⁽¹⁾	-	SDA	IOC	Y	-	-
RC5	24	39	43	43	AN17	_	_	OPA3IN0+	_	-	PRG3F ⁽¹⁾	T4IN ⁽¹⁾	_	_	_	_	MD2MOD ⁽¹⁾	_	_	IOC	Y	_	_
RC6	25	40	44	44	AN18	_	—	OPA3OUT OPA4IN1+ OPA4IN1-	C5IN1- C6IN1- C7IN1- C8IN1-		PRG3IN0 PRG4IN1	_	_	_	-	-	—		—	IOC	Y	_	_
RC7	26	1	1	1	AN19	_	_	OPA3IN0-	_		—	—	_	—		—	—		—	IOC	Y	—	—
RD0	19	34	38	38	AN20	_	_	OPA4IN0+		I	—	_	—	—		_	—		_	_	Y	—	_
RD1	20	35	39	39	AN21	_	_	OPA4OUT OPA3IN1+ OPA3IN1-	C1IN4- C2IN4- C3IN4- C4IN4- C5IN4- C6IN4- C7IN4- C8IN4-		PRG3IN1 PRG4IN0	_	_	_	_	_	_	_	_	_	Y	_	_
RD2	21	36	40	40	AN22	—	DAC4OUT1	OPA4IN0-	_	_	_	—	_	_	_	—	_	_	—	—	Y	_	_
RD3	22	37	41	41	AN23	—	_	_	C8IN2-	_	—	—	_	—		—	—		_	_	Y	_	—
RD4	27	2	2	2	AN24	—	_	_	C7IN2-		—	—	—	—	_	—	_	_	_	-	Y	—	-
RD5	28	3	3	3	AN25	—	-	—	C7IN3- C8IN3-		—	—	_	—		—	—		—	-	Y	_	-
RD6	29	4	4	4	AN26	_	—	_	C7IN1+	_	_	_	_	_	_	_	_	_	—	_	Y	_	_
RD7	30	5	5	5	AN27	_	_	_	C8IN1+		_	_	_	_	_	_		_	_		Y	_	-

40/44-PIN ALLOCATION TABLE (PIC16/L)E1777/0) (CONTINUED)

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Note 1:

Default peripheral input. Input can be moved to any other pin with the PPS input selection register. All pin outputs default to PORT latch data. Any pin can be selected as a digital peripheral output with the PPS output selection registers. These peripheral functions are bidirectional. The output pin selections must be the same as the input pin selections. 2: 3:

PIC16(L)F1777/8/9

		· ·	-	U/ - -	T-1 114	ALLOUA				,	1113) (C		NOLD,	,									
0/1	40-Pin PDIP	40-Pin (U)QFN	44-Pin TQFP	44-Pin QFN	ADC	VREF	DAC	op Amp	Comparator	ZCD	ÐЯЧ	Timers	PWM	ССР	900	CLC	Modulator	EUSART	MSSP	Interrupt	Sdn-Ilnd	High Current	Basic
RE0	8	23	25	25	AN5	DAC6REF1+ DAC8REF1+	-		-	-	-	—	-		—	_	-	_	-	Ι	Y	_	
RE1	9	24	26	26	AN6	DAC6REF1- DAC8REF1-	DAC6OUT1	_	-	_	_	_	_	_	_	—	_	_	-	_	Y	—	_
RE2	10	25	27	27	AN7	_	DAC8OUT1	-	_				_	_	_	_	-	_	_		Y	_	
RE3	1	16	18	18	—	-	—	_	-		-	-	-	_	-	-	-	-	-	IOC	Y	—	MCLR VPP
Vdd	11	7	7	7,8		_	_	-	_				_	_	_	_	-	_	_			_	VDD
Vdd	32	26	28	28		-	_	I	-	-		_	—	I	_	_	I	—	_	I	-	_	VDD
Vss	12	6	6	6		_	_	-	_				_	_	_	_	-	_	_			_	Vss
Vss	31	27	29	30		_	_	I	_	-		_	—	I	—	_		—	—		-	_	Vss
OUT ⁽²⁾		_			_	_	_	_	C1OUT C2OUT C3OUT C4OUT C5OUT C6OUT C7OUT C8OUT	_	_	_	PWM3 PWM4 PWM5 PWM9 PWM10 PWM11 PWM12	CCP1 CCP2 CCP7 CCP8	COG1A COG1B COG1D COG2A COG2B COG2C COG2D COG3A COG3C COG3B COG3C COG3D COG4A COG4B	CLC1OUT CLC2OUT CLC3OUT CLC4OUT	MD10UT MD20UT MD30UT MD40UT	DT ⁽³⁾ ТХ СК	SDO SDA(3) SCK SCL(3)	_	_	_	_

TABLE 4: 40/44-PIN ALLOCATION TABLE (PIC16(L)F1777/9) (CONTINUED)

1:

Default peripheral input. Input can be moved to any other pin with the PPS input selection register. All pin outputs default to PORT latch data. Any pin can be selected as a digital peripheral output with the PPS output selection registers. These peripheral functions are bidirectional. The output pin selections must be the same as the input pin selections. 2: 3:

Note

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1.0 DEVICE OVERVIEW

The PIC16(L)F1777/8/9 are described within this data sheet. See Table 2 for available package configurations.

Figure 1-1 shows a block diagram of the PIC16(L)F1777/8/9 devices. Table 1-2 shows the pinout descriptions.

Refer to Table 1-1 for peripherals available per device.

TABLE 1-1:DEVICE PERIPHERAL
SUMMARY

Peripheral		PIC16(L)F1778	PIC16(L)F1777/9
Analog-to-Digital Converter (ADC)		•	•
Fixed Voltage Reference (FVR)		•	•
Zero-Cross Detection (ZCD)		•	٠
Temperature Indicator		•	•
Complementary Output Generator (COG)			
	COG1	•	٠
	COG2	٠	•
	COG3	•	٠
	COG4		٠
Programmable Ramp Generator (PRG)			
	PRG1	•	•
	PRG2	•	•
	PRG3	٠	٠
	PRG4		•
10-bit Digital-to-Analog Converter (DAC)			
	DAC1	٠	•
	DAC2	٠	•
	DAC5	•	٠
	DAC6		٠
5-bit Digital-to-Analog Converter (DAC)			
	DAC3	٠	٠
	DAC4	•	٠
	DAC7	•	٠
	DAC8		•
Capture/Compare/PWM (CCP/ECCP) Mod	dules		-
	CCP1	•	٠
	CCP2	٠	٠
	CCP7	•	٠
	CCP8		٠
Comparators	1		
	C1	•	٠
	C2	٠	٠
	C3	•	٠
	C4	٠	٠
	C5	•	٠
	C6	٠	٠
	C7		٠
	C8		٠

TABLE 1-1: DEVICE PERIPHERAL SUMMARY

SUMMARY			
Peripheral		PIC16(L)F1778	PIC16(L)F1777/9
Configurable Logic Cell (CLC)			
	CLC1	٠	٠
	CLC2	•	٠
	CLC3	٠	٠
	CLC4	٠	٠
Data Signal Modulator (DSM)			
	DSM1	•	٠
	DSM2	•	٠
	DSM3	٠	٠
	DSM4		٠
Enhanced Universal Synchronous/Async Receiver/Transmitter (EUSART)	hronous		
	EUSART	٠	٠
Master Synchronous Serial Ports			
	MSSP	٠	•
Op Amps			
	OPA1	٠	٠
	OPA2	٠	•
	OPA3	٠	٠
	OPA4		•
10-bit Pulse-Width Modulator (PWM)			
	PWM3	٠	٠
	PWM4	٠	٠
	PWM9	٠	٠
	PWM10		٠
16-bit Pulse-Width Modulator (PWM)			
	PWM5	٠	٠
	PWM6	٠	٠
	PWM11	٠	٠
	PWM12		٠
8-bit Timers			
	Timer0	٠	٠
	Timer2	٠	٠
	Timer4	٠	٠
	Timer6	٠	٠
	Timer8	•	٠
16-bit Timers			
	Timer1	٠	٠
	Timer3	٠	٠
	Timer5	٠	٠

1.1 Register and Bit naming conventions

1.1.1 REGISTER NAMES

When there are multiple instances of the same peripheral in a device, the peripheral control registers will be depicted as the concatenation of a peripheral identifier, peripheral instance, and control identifier. The control registers section will show just one instance of all the register names with an 'x' in the place of the peripheral instance number. This naming convention may also be applied to peripherals when there is only one instance of that peripheral in the device to maintain compatibility with other devices in the family that contain more than one.

1.1.2 BIT NAMES

There are two variants for bit names:

- · Short name: Bit function abbreviation
- Long name: Peripheral abbreviation + short name

1.1.2.1 Short Bit Names

Short bit names are an abbreviation for the bit function. For example, some peripherals are enabled with the EN bit. The bit names shown in the registers are the short name variant.

Short bit names are useful when accessing bits in C programs. The general format for accessing bits by the short name is *RegisterName*bits. *ShortName*. For example, the enable bit, EN, in the COG1CON0 register can be set in C programs with the instruction COG1CON0bits.EN = 1.

Short names are generally not useful in assembly programs because the same name may be used by different peripherals in different bit positions. When this occurs, during the include file generation, all instances of that short bit name are appended with an underscore plus the name of the register in which the bit resides to avoid naming contentions.

1.1.2.2 Long Bit Names

Long bit names are constructed by adding a peripheral abbreviation prefix to the short name. The prefix is unique to the peripheral, thereby making every long bit name unique. The long bit name for the COG1 enable bit is the COG1 prefix, G1, appended with the enable bit short name, EN, resulting in the unique bit name G1EN.

Long bit names are useful in both C and assembly programs. For example, in C the COG1CON0 enable bit can be set with the GIEN = 1 instruction. In assembly, this bit can be set with the BSF COG1CON0, GIEN instruction.

1.1.2.3 Bit Fields

Bit fields are two or more adjacent bits in the same register. Bit fields adhere only to the short bit naming convention. For example, the three Least Significant bits of the COG1CON0 register contain the mode control bits. The short name for this field is MD. There is no long bit name variant. Bit field access is only possible in C programs. The following example demonstrates a C program instruction for setting the COG1 to the Push-Pull mode:

COG1CON0bits.MD = 0x5;

Individual bits in a bit field can also be accessed with long and short bit names. Each bit is the field name appended with the number of the bit position within the field. For example, the Most Significant mode bit has the short bit name MD2 and the long bit name is G1MD2. The following two examples demonstrate assembly program sequences for setting the COG1 to the Push-Pull mode:

EXAMPLE 1-1:

```
MOVLW ~(1<<G1MD1)
ANDWF COG1CON0,F
MOVLW 1<<G1MD2 | 1<<G1MD0
IORWF COG1CON0,F
```

EXAMPLE 1-2:

BSF COG1CON0,G1MD2 BCF COG1CON0,G1MD1 BSF COG1CON0,G1MD0

1.1.3 REGISTER AND BIT NAMING EXCEPTIONS

1.1.3.1 Status, Interrupt, and Mirror Bits

Status, interrupt enables, interrupt flags, and mirror bits are contained in registers that span more than one peripheral. In these cases, the bit name shown is unique so there is no prefix or short name variant.

1.1.3.2 Legacy Peripherals

There are some peripherals that do not strictly adhere to these naming conventions. Peripherals that have existed for many years and are present in almost every device are the exceptions. These exceptions were necessary to limit the adverse impact of the new conventions on legacy code. Peripherals that do adhere to the new convention will include a table in the registers section indicating the long name prefix for each peripheral instance. Peripherals that fall into the exception category will not have this table. These peripherals include, but are not limited to, the following:

- EUSART
- MSSP

PIC16(L)F1777/8/9

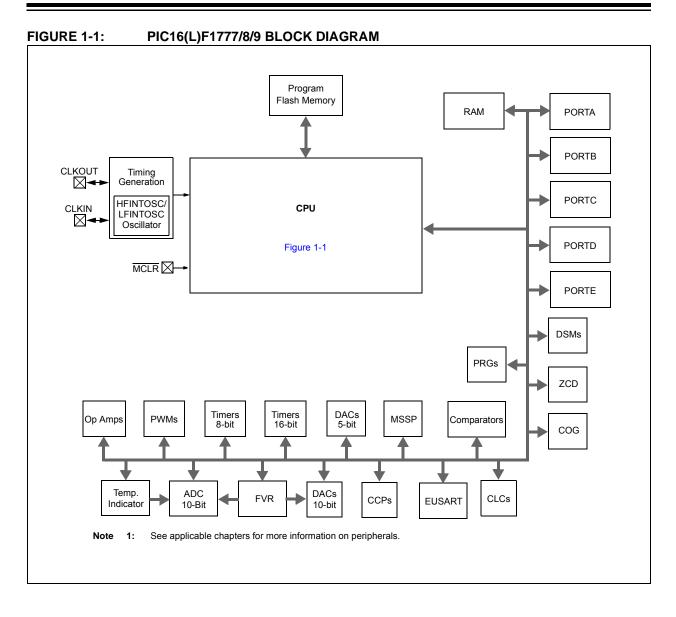


TABLE 1-2: PIC16(L)F1778 PINOUT DESCRIPTION

Name	Function	Input Type	Output Type	Description
RA0/AN0/C1IN0-/C2IN0-/	RA0	TTL/ST	CMOS	General purpose I/O.
C3IN0-/C4IN0-/C5IN0-/ C6IN0-/CLCIN0 ⁽¹⁾	AN0	AN		ADC Channel 0 input.
Cono-/CECINO.	C1IN0-	AN		Comparator 1 negative input.
	C2IN0-	AN		Comparator 2 negative input.
	C3IN0-	AN		Comparator 3 negative input.
	C4IN0-	AN	_	Comparator 4 negative input.
	C5IN0-	AN	_	Comparator 5 negative input.
	C6IN0-	AN		Comparator 6 negative input.
	CLCIN0 ⁽¹⁾	TTL/ST	_	CLC input 0.
RA1/AN1/C1IN1-/C2IN1-/	RA1	TTL/ST	CMOS	General purpose I/O.
C3IN1-/C4IN1-/PRG1IN0/	AN1	AN		Channel 1 input.
PRG2IN1/OPA1OUT/OPA2IN1+/ OPA2IN1-/CLCIN1 ⁽¹⁾	C1IN1-	AN	_	Comparator 1 negative input.
	C2IN1-	AN		Comparator 2 negative input.
	C3IN1-	AN		Comparator 3 negative input.
	C4IN1-	AN	_	Comparator 4 negative input.
	PRG1IN0	AN	_	Ramp generator 1 reference voltage input.
	PRG2IN1	AN		Ramp generator 2 reference voltage input.
	OPA1OUT	_	AN	Operational amplifier 1 output.
	OPA2IN1+	AN	_	Operational amplifier 2 non-inverting input.
	OPA2IN1-	AN	_	Operational amplifier 2 inverting input.
	CLCIN1 ⁽¹⁾	TTL/ST	_	CLC input 1.
RA2/AN2/VREF-/DAC1REF0-/	RA2	TTL/ST	CMOS	General purpose I/O.
DAC2REF0-/DAC3REF0-/	AN2	AN	_	ADC Channel 2 input.
DAC4REF0-/DAC5REF0-/ DAC7REF0-/C1IN0+/C2IN0+/	VREF-	AN	_	ADC negative reference.
C3IN0+/C4IN0+/C5IN0+/	DAC1REF0-	AN	_	DAC1 negative reference.
C6IN0+/DAC1OUT1	DAC2REF0-	AN	_	DAC2 negative reference.
	DAC3REF0-	AN	_	DAC3 negative reference.
	DAC4REF0-	AN	_	DAC4 negative reference.
	DAC5REF0-	AN	_	DAC5 negative reference.
	DAC7REF0-	AN	_	DAC7 negative reference.
	C1IN0+	AN		Comparator 1 positive input.
	C2IN0+	AN	—	Comparator 2 positive input.
	C3IN0+	AN		Comparator 3 positive input.
	C4IN0+	AN	—	Comparator 4 positive input.
	C5IN0+	AN	—	Comparator 5 positive input.
	C6IN0+	AN	—	Comparator 6 positive input.
	DAC1OUT1	—	AN	DAC1 voltage output.

 Legend:
 AN = Analog input or output
 CMOS = CMOS compatible input or output
 OD = Open-Drain

 TTL = TTL compatible input
 ST = Schmitt Trigger input with CMOS levels
 I²C = Schmitt Trigger input with I²C

 HP = High Power
 XTAL = Crystal levels

Note 1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.

2: All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.

Name	Function	Input Type	Output Type	Description
RA3/AN3/VREF+/DAC1REF0+/	RA3	TTL/ST	CMOS	General purpose I/O.
DAC2REF0+/DAC3REF0+/	AN3	AN		ADC Channel 3 input.
DAC4REF0+/DAC5REF0+/ DAC7REF0+/C1IN1+/MD1CL	VREF+	AN		ADC positive reference.
	DAC1REF0+	AN	_	DAC1 positive reference.
	DAC2REF0+	AN		DAC2 positive reference.
	DAC3REF0+	AN		DAC3 positive reference.
	DAC4REF0+	AN	_	DAC4 positive reference.
	DAC5REF0+	AN		DAC5 positive reference.
	DAC7REF0+	AN	—	DAC7 positive reference.
	C1IN1+	AN	—	Comparator 1 positive input.
	MD1CL ⁽¹⁾	TTL/ST	—	Data signal modulator 1 low carrier input.
RA4/OPA1IN0+/PRG1R/	RA4	TTL/ST	CMOS	General purpose I/O.
MD1CH/DAC4OUT1/T0CKI	OPA1IN0+	AN	—	Operational Amplifier 1 non-inverting input.
	PRG1R ⁽¹⁾	TTL/ST		Ramp generator set_rising input.
	MD1CH ⁽¹⁾	TTL/ST	—	Data signal modulator 1 high carrier input.
	DAC4OUT1	—	AN	DAC4 voltage output.
	T0CKI ⁽¹⁾	TTL/ST		Timer0 clock input.
RA5/AN4/OPA1IN0-/	RA5	TTL/ST	CMOS	General purpose I/O.
DAC2OUT1/PRG1F/	AN4	AN	_	ADC Channel 4 input.
MD1MOD/SS	OPA1IN0-	AN		Operational amplifier 1 inverting input.
	DAC2OUT1		AN	DAC2 voltage output.
	PRG1F ⁽¹⁾	TTL/ST	_	Ramp generator set_falling input.
	MD1MOD ⁽¹⁾	TTL/ST		Data signal modulator modulation input.
	SS	ST		Slave Select input.
RA6/CLKOUT/C6IN1+/OSC2	RA6	TTL/ST	CMOS	General purpose I/O.
	CLKOUT		CMOS	Fosc/4 output.
	C6IN1+	AN		Comparator 6 positive input.
	OSC2	XTAL	_	Crystal/Resonator (LP, XT, HS modes).
RA7/CLKIN/OSC1	RA7	TTL/ST	CMOS	General purpose I/O.
	CLKIN	TTL/ST	_	CLC input.
	OSC1	XTAL	_	Crystal/Resonator (LP, XT, HS modes).
RB0/AN12/ZCD/HIB0/C2IN1+/	RB0	TTL/ST	CMOS	General purpose I/O.
COG1IN	AN12	AN	—	ADC Channel 12 input.
	ZCD	AN	—	Zero-cross detection input.
	HIB0	HP	HP	High-Power output.
	C2IN1+	AN	_	Comparator 2 positive input.
	COG1IN ⁽¹⁾	TTL/ST	_	Complementary output generator 1 input.

TABLE 1-2: PIC16(L)F1778 PINOUT DESCRIPTION (CONTINUED)

Legend:AN = Analog input or outputCMOS = CMOS compatible input or outputOD= Open-DrainTTL = TTL compatible inputST= Schmitt Trigger input with CMOS levelsI²C= Schmitt Trigger input with I²CHP = High PowerXTAL= Crystal levels

Note 1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.
 All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.

TABLE 1-2: PIC16(L)F1778 PINOUT DESCRIPTION (CONTINUED)

Name	Function	Input Type	Output Type	Description
RB1/AN10/PRG1IN1/PRG2IN0/	RB1	TTL/ST	CMOS	General purpose I/O.
HIB1/C1IN3-/C2IN3-/C3IN3-/	AN10	AN	_	ADC Channel 10 input.
C4IN3-/OPA2OUT/OPA1IN1+/ OPA1IN1-/COG2IN	PRG1IN1	AN	_	Ramp generator 1 reference voltage input.
	PRG2IN0	AN	_	Ramp generator 2 reference voltage input.
	HIB1	HP	HP	High-Power output.
	C1IN3-	AN	_	Comparator 1 negative input.
	C2IN3-	AN	_	Comparator 2 negative input.
	C3IN3-	AN	_	Comparator 3 negative input.
	C4IN3-	AN	_	Comparator 4 negative input.
	OPA2OUT		AN	Operational amplifier 2 output.
	OPA1IN1+	AN		Operational amplifier 1 non-inverting input.
	OPA1IN1-	AN	_	Operational amplifier 1 inverting input.
	COG2IN ⁽¹⁾	TTL/ST	_	Complementary output generator 2 input.
RB2/AN8/OPA2IN0-/	RB2	TTL/ST	CMOS	General purpose I/O.
DAC3OUT1/COG3IN	AN8	AN	_	ADC Channel 8 input.
	OPA2IN0-	AN	_	Operational amplifier 2 inverting input.
	DAC3OUT1	_	AN	DAC3 voltage output.
	COG3IN ⁽¹⁾	TTL/ST	_	Complementary output generator 3 input.
RB3/AN9/C1IN2-/C2IN2-/	RB3	TTL/ST	CMOS	General purpose I/O.
C3IN2-/OPA2IN0+/MD3CL	AN9	AN	_	ADC Channel 9 input.
	C1IN2-	AN	_	Comparator 1 negative input.
	C2IN2-	AN		Comparator 2 negative input.
	C3IN2-	AN	_	Comparator 3 negative input.
	OPA2IN0+	AN		Operational amplifier 2 non-inverting input.
	MD3CL ⁽¹⁾	TTL/ST		Data signal modulator 3 low carrier input.
RB4/AN11/C3IN1+/T5G/MD3CH	RB4	TTL/ST	CMOS	General purpose I/O.
	AN11	AN		ADC Channel 11 input.
	C3IN1+	AN		Comparator 3 positive input.
	T5G ⁽¹⁾	TTL/ST		Timer5 gate input.
	MD3CH ⁽¹⁾	TTL/ST		Data signal modulator 3 high carrier input.
RB5/AN13/DAC5REF1-/	RB5	TTL/ST	CMOS	General purpose I/O.
DAC7REF1-/C4IN2-/T1G/CCP7/	AN13	AN	—	ADC Channel 11 input.
MD3MOD	DAC5REF1-	AN	_	DAC5 negative reference.
	DAC7REF1-	AN	_	DAC7 negative reference.
	C4IN2-	AN	_	Comparator 4 negative input.
	T1G ⁽¹⁾	TTL/ST	_	Timer1 gate input.
	CCP7 ⁽¹⁾	TTL/ST	_	CCP7 capture input.

 Legend:
 AN = Analog input or output
 CMOS = CMOS compatible input or output
 OD = Open-Drain

 TTL = TTL compatible input
 ST = Schmitt Trigger input with CMOS levels
 I²C = Schmitt Trigger input with I²C

 HP = High Power
 XTAL = Crystal levels

Note 1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.
 All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.

		Input	Output	DN (CONTINUED)
Name	Function	Туре	Туре	Description
RB6/DAC5REF1+/DAC7REF1+/	RB6	TTL/ST	CMOS	General purpose I/O.
C4IN1+/CLCIN2/ICSPCLK	DAC5REF1+	AN	_	DAC5 positive reference.
	DAC7REF1+	AN	_	DAC7 positive reference.
	C4IN1+	AN	_	Comparator 2 positive input.
	CLCIN2 ⁽¹⁾	TTL/ST	_	CLC input 2.
	ICSPCLK	ST	_	Serial Programming Clock.
RB7/C5IN1+/DAC1OUT2/	RB7	TTL/ST	CMOS	General purpose I/O.
	C5IN1+	AN		Comparator 5 positive input.
DAC4OUT2/DAC5OUT2/ DAC7OUT2/T6IN/CLCIN3/	DAC10UT2	—	AN	DAC1 voltage output.
ICSPDAT	DAC2OUT2	—	AN	DAC2 voltage output.
	DAC3OUT2	—	AN	DAC3 voltage output.
	DAC4OUT2	_	AN	DAC4 voltage output.
	DAC5OUT2	—	AN	DAC5 voltage output.
	DAC7OUT2	—	AN	DAC7 voltage output.
	T6IN ⁽¹⁾	TTL/ST		Timer6 gate input.
	CLCIN3 ⁽¹⁾	TTL/ST		CLC input 3.
	ICSPDAT	ST	CMOS	ICSP™ Data I/O.
RC0/DAC5OUT1/T1CKI/T3CKI/	RC0	TTL/ST	CMOS	General purpose I/O.
T3G/SOSCO	DAC5OUT1	—	AN	DAC5 voltage output.
	T1CKI ⁽¹⁾	AN	_	Comparator 4 negative input.
	T3CKI ⁽¹⁾	TTL/ST		Timer3 clock input.
	T3G ⁽¹⁾	TTL/ST		Timer3 gate input.
	SOSCO	—	XTAL	Secondary oscillator output.
RC1/DAC7OUT1/PRG2R/CCP2/	RC1	TTL/ST	CMOS	General purpose I/O.
SOSCI	DAC7OUT1	_	AN	DAC7 voltage output.
	PRG2R ⁽¹⁾	TTL/ST		Ramp generator set_rising input.
	CCP2 ⁽¹⁾	TTL/ST	_	CCP2 capture input.
	SOSCI	XTAL		Secondary oscillator input.
RC2/AN14/C5IN2-/C6IN2-/	RC2	TTL/ST	CMOS	General purpose I/O.
PRG2F/CCP1/T5CKI	AN14	AN		ADC Channel 14 input.
	C5IN2-	AN	_	Comparator 5 negative input.
	C6IN2-	AN		Comparator 6 negative input.
	PRG2F ⁽¹⁾	TTL/ST		Ramp generator set_falling input.
	CCP1 ⁽¹⁾	TTL/ST	_	CCP1 capture input.
	T5CKI ⁽¹⁾	TTL/ST	-	Timer5 clock input.

TABLE 1-2:	PIC16(L)F1778 PINOUT DESCRIPTION (CONTINUED	1
			,

Legend: AN = Analog input or output CMOS = CMOS compatible input or output

= Open-Drain

OD TTL = TTL compatible input ST = Schmitt Trigger input with CMOS levels I^2C = Schmitt Trigger input with I^2C HP = High Power XTAL = Crystal levels

Note 1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.

2: All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.

TABLE 1-2: PIC16(L)F1778 PINOUT DESCRIPTION (CONTINUED)

Name	Function	Input Type	Output Type	Description
RC3/AN15/C1IN4-/C2IN4-/	RC3	TTL/ST	CMOS	General purpose I/O.
C3IN4-/C4IN4-/C5IN4-/C6IN4-/	AN15	AN	_	ADC Channel 15 input.
T2IN//MD2CL/SCL	C1IN4-	AN	_	Comparator 1 negative input.
	C2IN4-	AN	_	Comparator 2 negative input.
	C3IN4-	AN		Comparator 3 negative input.
	C4IN4-	AN	_	Comparator 4 negative input.
	C5IN4-	AN		Comparator 5 negative input.
	C6IN4-	AN	_	Comparator 6 negative input.
	T2IN ⁽¹⁾	TTL/ST	_	Timer2 gate input.
	MD2CL ⁽¹⁾	TTL/ST	_	Data signal modulator 2 low carrier input.
	SCL	l ² C	OD	I ² C clock.
RC4/AN16/C5IN3-/C6IN3-/T8IN/	RC4	TTL/ST	CMOS	General purpose I/O.
PRG3R/MD2CH/SDA	AN16	AN		ADC Channel 16 input.
	C5IN3-	AN	_	Comparator 5 negative input.
	C6IN3-	AN	_	Comparator 6 negative input.
	T8IN ⁽¹⁾	TTL/ST	_	Timer8 gate input.
	PRG3R ⁽¹⁾	TTL/ST	_	Ramp generator set_rising input.
	MD2CH ⁽¹⁾	TTL/ST	_	Data signal modulator 2 high carrier input.
	SDA	l ² C	OD	I ² C data input/output.
RC5/AN17/OPA3IN0+/T4IN/	RC5	TTL/ST	CMOS	General purpose I/O.
PRG3F/MD2MOD	AN17	AN	_	ADC Channel 17 input.
	OPA3IN0+	AN	_	Operational amplifier 3 inverting input.
	T4IN ⁽¹⁾	TTL/ST	_	Timer4 gate input.
	PRG3F ⁽¹⁾	TTL/ST	_	Ramp generator set_falling input.
	MD2MOD ⁽¹⁾	TTL/ST	_	Data signal modulator modulation input.
RC6/AN18/PRG3IN0/C5IN1-/	RC6	TTL/ST	CMOS	General purpose I/O.
C6IN1-/OPA3OUT	AN18	AN	_	ADC Channel 18 input.
	PRG3IN0	AN		Ramp generator 3 reference voltage input.
	C5IN1-	AN	_	Comparator 5 negative input.
	C6IN1-	AN		Comparator 6 negative input.
	OPA3OUT	_	AN	Operational amplifier 3 output.
RC7/AN19/OPA3IN0-	RC7	TTL/ST	CMOS	General purpose I/O.
	AN19	AN	—	ADC Channel 19 input.
	OPA3IN0-	AN	—	Operational amplifier 3 non-inverting input.
RE3/MCLR	RE3	TTL/ST	CMOS	General purpose input.
	MCLR	ST	—	Master clear input.
Vdd	Vdd	Power	—	Positive supply.
Vss	Vss	Power		Ground reference.

 Legend:
 AN = Analog input or output
 CMOS = CMOS compatible input or output
 OD = Open-Drain

 TTL = TTL compatible input
 ST = Schmitt Trigger input with CMOS levels
 I²C = Schmitt Trigger input with I²C

 HP = High Power
 XTAL = Crystal levels

Note 1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.
 All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.

Name	Function	Input Type	Output Type	Description
2)	C1OUT		CMOS	Comparator 1 output.
	C2OUT		CMOS	Comparator 2 output.
	C3OUT		CMOS	Comparator 3 output.
	C4OUT		CMOS	Comparator 4 output.
	C5OUT		CMOS	Comparator 5 output.
	C6OUT		CMOS	Comparator 6 output.
	CCP1		CMOS	Compare/PWM1 output.
	CCP2		CMOS	Compare/PWM2 output.
	CCP7		CMOS	Compare/PWM7 output.
	MD10UT		CMOS	Data signal modulator 1 output.
	MD2OUT		CMOS	Data signal modulator 2 output.
	MD3OUT		CMOS	Data signal modulator 3 output.
	PWM3OUT		CMOS	PWM3 output.
	PWM4OUT		CMOS	PWM4 output.
	PWM5OUT		CMOS	PWM5 output.
	PWM6OUT		CMOS	PWM6 output.
	PWM9OUT		CMOS	PWM9 output.
	PWM11OUT		CMOS	PWM11 output.
	COG1A		CMOS	Complementary output generator 1 output A.
	COG1B		CMOS	Complementary output generator 1 output B.
	COG1C		CMOS	Complementary output generator 1 output C.
	COG1D		CMOS	Complementary output generator 1 output D.
	COG2A		CMOS	Complementary output generator 2 output A.
	COG2B		CMOS	Complementary output generator 2 output B.
	COG2C		CMOS	Complementary output generator 2 output C.
	COG2D		CMOS	Complementary output generator 2 output D.
	COG3A		CMOS	Complementary output generator 3 output A.
	COG3B		CMOS	Complementary output generator 3 output B.
	COG3C		CMOS	Complementary output generator 3 output C.
	COG3D		CMOS	Complementary output generator 3 output D.
	SDA ⁽³⁾		OD	I ² C data output.
	SCK		CMOS	SPI clock output.
	SCL ⁽³⁾		OD	I ² C clock output.
	SDO		CMOS	SPI data output.
	ТХ		CMOS	EUSART asynchronous TX data out.
	CK		CMOS	EUSART synchronous clock out.
	DT ⁽³⁾		CMOS	EUSART synchronous data output.
	CLC1OUT		CMOS	Configurable logic cell 1 output.
	CLC2OUT		CMOS	Configurable logic cell 2 output.
	CLC3OUT		CMOS	Configurable logic cell 3 output.
	CLC4OUT		CMOS	Configurable logic cell 4 output.

TABLE 1-2: PIC16(L)F1778 PINOUT DESCRIPTION (CONTINUED)

TTL = TTL compatible input ST = Schmitt Trigger input with CMOS levels I^2C = Schmitt Trigger input with I^2C HP = High Power XTAL = Crystal levels

Note 1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.

2: All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.

TABLE 1-3: PIC16(L)F1777/9 PINOUT DESCRIPTION

Name	Function	Input Type	Output Type	Description
RA0/AN0/C1IN0-/C2IN0-/	RA0	TTL/ST	CMOS	General purpose I/O.
C3IN0-/C4IN0-/C5IN0-/ C6IN0-/C7IN0-/C8IN0-/CLCIN0	AN0	AN		ADC Channel 0 input.
	C1IN0-	AN	_	Comparator 1 negative input.
	C2IN0-	AN	_	Comparator 2 negative input.
	C3IN0-	AN	_	Comparator 3 negative input.
	C4IN0-	AN		Comparator 4 negative input.
	C5IN0-	AN		Comparator 5 negative input.
	C6IN0-	AN		Comparator 6 negative input.
	C7IN0-	AN		Comparator 7 negative input.
	C8IN0-	AN	_	Comparator 8 negative input.
	CLCIN0 ⁽¹⁾	TTL/ST		CLC input 0.
RA1/AN1/C1IN1-/C2IN1-/	RA1	TTL/ST	CMOS	General purpose I/O.
C3IN1-/C4IN1-/PRG1IN0/ PRG2IN1/OPA1OUT/OPA2IN1+/	AN1	AN		Channel 1 input.
OPA2IN1/OPA1001/OPA2IN1+/	C1IN1-	AN		Comparator 1 negative input.
	C2IN1-	AN	_	Comparator 2 negative input.
	C3IN1-	AN		Comparator 3 negative input.
	C4IN1-	AN		Comparator 4 negative input.
	PRG1IN0	AN	_	Ramp generator 1 reference voltage input.
	PRG2IN1	AN	_	Ramp generator 2 reference voltage input.
	OPA10UT	—	AN	Operational amplifier 1 output.
	OPA2IN1+	AN		Operational amplifier 2 non-inverting input.
	OPA2IN1-	AN		Operational amplifier 2 inverting input.
	CLCIN1 ⁽¹⁾	TTL/ST		CLC input 0.

Legend: AN = Analog input or output CMOS = CMOS compatible input or output OD = Open-Drain

TTL = TTL compatible input ST = Schmitt Trigger input with CMOS levels I^2C = Schmitt Trigger input with I^2C HP = High Power XTAL = Crystal levels

Note 1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.
 All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.

Name	Function	Input Type	Output Type	Description
RA2/AN2/DAC1REF0-/	RA2	TTL/ST	CMOS	General purpose I/O.
DAC2REF0-/DAC3REF0-/	AN2	AN	_	ADC Channel 2 input.
DAC4REF0-/DAC5REF0-/ DAC6REF0-/DAC7REF0-/	DAC1REF0-	AN	_	DAC1 negative reference.
DAC8REF0-/C1IN0+/C2IN0+/	DAC2REF0-	AN	_	DAC2 negative reference.
C3IN0+/C4IN0+/C5IN0+/	DAC3REF0-	AN	_	DAC3 negative reference.
C6IN0+/C7IN0+/C8IN0+/ DAC1OUT1	DAC4REF0-	AN	_	DAC4 negative reference.
	DAC5REF0-	AN		DAC5 negative reference.
	DAC6REF0-	AN		DAC6 negative reference.
	DAC7REF0-	AN	_	DAC7 negative reference.
	DAC8REF0-	AN		DAC8 negative reference.
	C1IN0+	AN	_	Comparator 1 positive input.
	C2IN0+	AN	_	Comparator 2 positive input.
	C3IN0+	AN	_	Comparator 3 positive input.
	C4IN0+	AN	_	Comparator 4 positive input.
	C5IN0+	AN	_	Comparator 5 positive input.
	C6IN0+	AN	_	Comparator 6 positive input.
	C7IN0+	AN	_	Comparator 7 positive input.
	C8IN0+	AN	_	Comparator 8 positive input.
	DAC1OUT1	_	AN	DAC1 voltage output.
RA3/AN3/DAC1REF0+/	RA3	TTL/ST	CMOS	General purpose I/O.
DAC2REF0+/DAC3REF0+/	AN3	AN	_	ADC Channel 3 input.
DAC4REF0+/DAC5REF0+/ DAC6REF0+/DAC7REF0+/	VREF+	AN		ADC positive reference.
DAC8REF0+/C1IN1+/MD1CL	DAC1REF0+	AN	_	DAC1 positive reference.
	DAC2REF0+	AN	_	DAC2 positive reference.
	DAC3REF0+	AN		DAC3 positive reference.
	DAC4REF0+	AN	_	DAC4 positive reference.
	DAC5REF0+	AN	_	DAC5 positive reference.
	DAC6REF0+	AN		DAC6 positive reference.
	DAC7REF0+	AN		DAC7 positive reference.
	DAC8REF0+	AN	_	DAC8 positive reference.
	C1IN1+	AN		Comparator 1 positive input.
	MD1CL ⁽¹⁾	TTL/ST		Data signal modulator 1 low carrier input.
RA4/OPA1IN0+/PRG1R/MD1CH	RA4	TTL/ST	CMOS	General purpose I/O.
	OPA1IN0+	AN	_	Operational Amplifier 1 non-inverting input.
	PRG1R ⁽¹⁾	TTL/ST	—	Ramp generator set_rising input.
	MD1CH ⁽¹⁾	TTL/ST		Data signal modulator 1 high carrier input.

TABLE 1-3: PIC16(L)F1777/9 PINOUT DESCRIPTION (CONTINUED)

 Legend:
 AN = Analog input or output
 CMOS = CMOS compatible input or output
 OD = Open-Drain

 TTL = TTL compatible input
 ST = Schmitt Trigger input with CMOS levels
 I²C = Schmitt Trigger input with I²C

 HP = High Power
 XTAL = Crystal levels

Note 1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.

2: All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.

TABLE 1-3: PIC16(L)F1777/9 PINOUT DESCRIPTION (CONTINUED)

Name	Function	Input Type	Output Type	Description
RA5/AN4/OPA1IN0-/	RA5	TTL/ST	CMOS	General purpose I/O.
DAC2OUT1/PRG1F/	AN4	AN	_	ADC Channel 4 input.
MD1MOD/SS	OPA1IN0-	AN	_	Operational amplifier 1 inverting input.
	DAC2OUT1	_	AN	DAC2 voltage output.
	PRG1F ⁽¹⁾	TTL/ST	_	Ramp generator set_falling input.
	MD1MOD ⁽¹⁾	TTL/ST	_	Data signal modulator modulation input.
	SS	ST		Slave Select input.
RA6/CLKOUT/C6IN1+/OSC2	RA6	TTL/ST	CMOS	General purpose I/O.
	CLKOUT		CMOS	Fosc/4 output.
	C6IN1+	AN	_	Comparator 6 positive input.
	OSC2	XTAL	_	Crystal/Resonator (LP, XT, HS modes).
RA7/CLKIN/OSC1	RA7	TTL/ST	CMOS	General purpose I/O.
	CLKIN	TTL/ST	_	CLC input.
	OSC1	XTAL		Crystal/Resonator (LP, XT, HS modes).
RB0/AN12/ZCD/HIB0/C2IN1+/	RB0	TTL/ST	CMOS	General purpose I/O.
CCP8/COG1IN/MD4CL/	AN12	AN	—	ADC Channel 12 input.
INT	ZCD	AN	_	Zero-cross detection input.
	HIB0	HP	HP	High-Power output.
	C2IN1+	AN	_	Comparator 2 positive input.
	CCP8 ⁽¹⁾	TTL/ST	_	CCP8 capture input.
	COG1IN ⁽¹⁾	TTL/ST	_	Complementary output generator 1 input.
	MD4CL ⁽¹⁾	TTL/ST	_	Data signal modulator 4 low carrier input.
	INT	TTL/ST		External interrupt.
RB1/AN10/PRG1IN1/PRG2IN0/	RB1	TTL/ST	CMOS	General purpose I/O.
PRG4R/HIB1/C1IN3-/C2IN3-/	AN10	AN	_	ADC Channel 10 input.
C3IN3-/C4IN3-/OPA2OUT/ OPA1IN1+/OPA1IN1-/COG2IN/	PRG1IN1	AN	_	Ramp generator 1 reference voltage input.
MD4CH	PRG2IN0	AN	_	Ramp generator 2 reference voltage input.
	PRG4R ⁽¹⁾	TTL/ST		Ramp generator set_rising input.
	HIB1	HP	HP	High-Power output.
	C1IN3-	AN	_	Comparator 1 negative input.
	C2IN3-	AN		Comparator 2 negative input.
	C3IN3-	AN	_	Comparator 3 negative input.
	C4IN3-	AN		Comparator 4 negative input.
	OPA2OUT		AN	Operational amplifier 2 output.
	OPA1IN1+	AN		Operational amplifier 1 non-inverting input.
	OPA1IN1-	AN	_	Operational amplifier 1 inverting input.
	COG2IN ⁽¹⁾	TTL/ST	—	Complementary output generator 2 input.
	MD4CH ⁽¹⁾	TTL/ST		Data signal modulator 4 high carrier input.

Legend:AN = Analog input or outputCMOS = CMOS compatible input or outputOD= Open-DrainTTL = TTL compatible inputST= Schmitt Trigger input with CMOS levelsI²C= Schmitt Trigger input with I²CHP = High PowerXTAL = Crystal levelsXTAL = Crystal levelsI

Note 1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.
 All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.

Name	Function	Input Type	Output Type	Description
RB2/AN8/OPA2IN0-/	RB2	TTL/ST	CMOS	General purpose I/O.
DAC3OUT1/PRG4F/COG3IN/	AN8	AN		ADC Channel 8 input.
MD4MOD	OPA2IN0-	AN	_	Operational amplifier 2 inverting input.
	DAC3OUT1	—	AN	DAC3 voltage output.
	PRG4F ⁽¹⁾	TTL/ST		Ramp generator set_falling input.
	COG3IN ⁽¹⁾	TTL/ST	_	Complementary output generator 3 input.
	MD4MOD ⁽¹⁾	TTL/ST	_	Data signal modulator modulation input.
RB3/AN9/C1IN2-/C2IN2-/	RB3	TTL/ST	CMOS	General purpose I/O.
C3IN2-/OPA2IN0+/MD3CL	AN9	AN	_	ADC Channel 9 input.
	C1IN2-	AN	_	Comparator 1 negative input.
	C2IN2-	AN	_	Comparator 2 negative input.
	C3IN2-	AN	_	Comparator 3 negative input.
	OPA2IN0+	AN		Operational amplifier 2 non-inverting input.
	MD3CL ⁽¹⁾	TTL/ST	_	Data signal modulator 3 low carrier input.
RB4/AN11/C3IN1+/MD3CH	RB4	TTL/ST	CMOS	General purpose I/O.
	AN11	AN	_	ADC Channel 11 input.
	C3IN1+	AN	_	Comparator 3 positive input.
	MD3CH ⁽¹⁾	TTL/ST	_	Data signal modulator 3 high carrier input.
RB5/AN13/DAC5REF1-/	RB5	TTL/ST	CMOS	General purpose I/O.
DAC7REF1-/C4IN2-/CCP7/	AN13	AN	_	ADC Channel 11 input.
MD3MOD	DAC5REF1-	AN	_	DAC5 negative reference.
	DAC7REF1-	AN	_	DAC7 negative reference.
	C4IN2-	AN	_	Comparator 4 negative input.
	CCP7 ⁽¹⁾	TTL/ST	_	CCP7 capture input.
	MD3MOD ⁽¹⁾	TTL/ST	_	Data signal modulator modulation input.
RB6/DAC5REF1+/DAC7REF1+/	RB6	TTL/ST	CMOS	General purpose I/O.
C4IN1+/CLCIN2/ICSPCLK	DAC5REF1+	AN	_	DAC5 positive reference.
	DAC7REF1+	AN	—	DAC7 positive reference.
	C4IN1+	AN	_	Comparator 2 positive input.
	CLCIN2 ⁽¹⁾	TTL/ST		CLC input 2.
	ICSPCLK	ST	_	Serial Programming Clock.

TABLE 1-3: PIC16(L)F1777/9 PINOUT DESCRIPTION (CONTINUED)

Legend:AN = Analog input or outputCMOS = CMOS compatible input or outputOD= Open-DrainTTL = TTL compatible inputST= Schmitt Trigger input with CMOS levelsI²C= Schmitt Trigger input with I²CHP = High PowerXTAL= Crystal levels

Note 1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.

2: All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.

TABLE 1-3: PIC16(L)F1777/9 PINOUT DESCRIPTION (CONTINUED)

Name	Function	Input Type	Output Type	Description					
RB7/C5IN1+/DAC1OUT2/	RB7	TTL/ST	CMOS	General purpose I/O.					
DAC2OUT2/DAC3OUT2/	C5IN1+	AN		Comparator 5 positive input.					
DAC4OUT2/DAC5OUT2/ DAC6OUT2/DAC7OUT2/	DAC1OUT2		AN	DAC1 voltage output.					
DAC8OUT2/T6IN/CLCIN3/	DAC2OUT2	_	AN	DAC2 voltage output.					
CSPDAT	DAC3OUT2		AN	DAC3 voltage output.					
	DAC4OUT2		AN	DAC4 voltage output.					
	DAC5OUT2	—	AN	DAC5 voltage output.					
	DAC6OUT2		AN	DAC6 voltage output.					
	DAC7OUT2		AN	DAC7 voltage output.					
	DAC8OUT2	_	AN	DAC8 voltage output.					
	T6IN ⁽¹⁾	TTL/ST	_	Timer6 gate input.					
	CLCIN3 ⁽¹⁾	TTL/ST	_	CLC input 3.					
	ICSPDAT	ST	CMOS	ICSP™ Data I/O.					
RC0/DAC5OUT1/T1CKI/T3CKI/	RC0	TTL/ST	CMOS	General purpose I/O.					
T3G/SOSCO	DAC5OUT1	—	AN	DAC5 voltage output.					
	T1CKI ⁽¹⁾	AN		Comparator 4 negative input.					
	T3CKI ⁽¹⁾	TTL/ST		Timer3 clock input.					
	T3G ⁽¹⁾	TTL/ST	_	Timer3 gate input.					
	SOSCO	—	XTAL	Secondary oscillator output.					
RC1/DAC7OUT1/PRG2R/CCP2/	RC1	TTL/ST	CMOS	General purpose I/O.					
SOSCI	DAC7OUT1	_	AN	DAC7 voltage output.					
	PRG2R ⁽¹⁾	TTL/ST		Ramp generator set_rising input.					
	CCP2 ⁽¹⁾	TTL/ST		CCP2 capture input.					
	SOSCI	XTAL		Secondary oscillator input.					
RC2/AN14/C5IN2-/C6IN2-/	RC2	TTL/ST	CMOS	General purpose I/O.					
PRG2F/CCP1	AN14	AN		ADC Channel 14 input.					
	C5IN2-	AN		Comparator 5 negative input.					
	C6IN2-	AN		Comparator 6 negative input.					
RC3/AN15/C1IN4-/C2IN4-/	RC3	TTL/ST	CMOS	General purpose I/O.					
C3IN4-/C4IN4-/C5IN4-/C6IN4-/	AN15	AN		ADC Channel 15 input.					
C7IN4-/C8IN4-/T2IN/MD2CL/ SCL	C1IN4-	AN		Comparator 1 negative input.					
SCL	C2IN4-	AN		Comparator 2 negative input.					
	C3IN4-	AN	_	Comparator 3 negative input.					
	C4IN4-	AN		Comparator 4 negative input.					
	C5IN4-	AN		Comparator 5 negative input.					
	C6IN4-	AN		Comparator 6 negative input.					
	C7IN4-	AN	—	Comparator 7 negative input.					
	C8IN4-	AN		Comparator 8 negative input.					
	T2IN ⁽¹⁾	TTL/ST		Timer2 gate input.					
	MD2CL ⁽¹⁾	TTL/ST	_	Data signal modulator 2 low carrier input.					
	SCL	l ² C	OD	I ² C clock.					

Note 1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.

2: All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.

Name	Function	Input Type	Output Type	Description					
RC4/AN16/C5IN3-/C6IN3-/	RC4	TTL/ST	CMOS	General purpose I/O.					
PRG3R/T8IN/MD2CH/SDA	AN16	AN	_	ADC Channel 16 input.					
	C5IN3-	AN	_	Comparator 5 negative input.					
	C6IN3-	AN	_	Comparator 6 negative input.					
	PRG3R ⁽¹⁾	TTL/ST	_	Ramp generator set_rising input.					
	T8IN ⁽¹⁾	TTL/ST	_	Timer8 gate input.					
	MD2CH ⁽¹⁾	TTL/ST	_	Data signal modulator 2 high carrier input.					
	SDA	l ² C	OD	I ² C data input/output.					
RC5/AN17/OPA3IN0+/PRG3F/	RC5	TTL/ST	CMOS	General purpose I/O.					
T4IN/MD2MOD	AN17	AN	_	ADC Channel 17 input.					
	OPA3IN0	AN	_	Operational amplifier 3 inverting input.					
	PRG3F ⁽¹⁾	TTL/ST	_	Ramp generator set_falling input.					
	T4IN ⁽¹⁾	TTL/ST	_	Timer4 gate input.					
	MD2MOD ⁽¹⁾	TTL/ST	_	Data signal modulator modulation input.					
RC6/AN18/PRG3IN0/PRG4IN1/	RC6	TTL/ST	CMOS	General purpose I/O.					
C5IN1-/C6IN1-/C7IN1-/C8IN1-/	AN18	AN	_	ADC Channel 18 input.					
OPA3OUT/OPA4IN1+/OPA4IN1-	PRG3IN0	PRG3IN0 AN — Ramp generator 3 reference		Ramp generator 3 reference voltage input.					
	PRG4IN1	AN	_	Ramp generator 4 reference voltage input.					
	C5IN1-	AN	_	Comparator 5 negative input.					
	C6IN1-	AN	_	Comparator 6 negative input.					
	C7IN1-	C7IN1- AN — Comparator 7 negative input.							
	C8IN1-	AN	_	Comparator 8 negative input.					
	OPA3OUT	_	AN	Operational amplifier 3 output.					
	OPA4IN1+	AN	_	Operational amplifier 4 non-inverting input.					
	OPA4IN1-	AN		Operational amplifier 4 inverting input.					
RC7/AN19/OPA3IN0-	RC7	TTL/ST	CMOS	General purpose I/O.					
	AN19	AN	_	ADC Channel 19 input.					
	OPA3IN0-	AN	—	Operational amplifier 3 non-inverting input.					
RD0/AN20/OPA4IN0+	RD0	TTL/ST	CMOS	General purpose I/O.					
	AN20	AN	—	ADC Channel 20 input.					
	OPA4IN0+	AN	—	Operational amplifier 4 non-inverting input.					

TABLE 1-3:	PIC16(L)F1777/9 PINOUT DESCRIPTION (CONTINUED)
IADEE 1-3.	

Legend: AN = Analog input or output CMOS = CMOS compatible input or output = Open-Drain OD TTL = TTL compatible input ST = Schmitt Trigger input with CMOS levels I^2C = Schmitt Trigger input with I^2C HP = High Power

XTAL = Crystal levels

Note 1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.

2: All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.

TABLE 1-3: PIC16(L)F1777/9 PINOUT DESCRIPTION (CONTINUED)

Name	Function	Input Type	Output Type	Description					
RD1/AN21/PRG3IN1/PRG4IN0/	RD1	TTL/ST	CMOS	General purpose I/O.					
C1IN4-/C2IN4-/C3IN4-/C4IN4-/	AN21	AN21 AN — ADC Channel 21 input.							
C5IN4-/C6IN4-/C7IN4-/C8IN4-/ OPA4OUT/OPA3IN1+/OPA3IN1-	PRG3IN1	AN		Ramp generator 3 reference voltage input.					
	PRG4IN0	AN		Ramp generator 4 reference voltage input.					
	C1IN4-	AN	_	Comparator 1 negative input.					
	C2IN4-	AN		Comparator 2 negative input.					
	C3IN4-	AN		Comparator 3 negative input.					
	C4IN4-	AN		Comparator 4 negative input.					
	C5IN4-	AN		Comparator 5 negative input.					
	C6IN4-	AN		Comparator 6 negative input.					
	C7IN4-	AN		Comparator 7 negative input.					
	C8IN4-	AN		Comparator 8 negative input.					
	OPA4OUT	_	AN	Operational amplifier 4 output.					
	OPA3IN1+	AN		Operational amplifier 3 non-inverting input.					
	OPA3IN1-	AN		Operational amplifier 3 inverting input.					
RD2/AN22/DAC4OUT1/	RD2	TTL/ST	CMOS	General purpose I/O.					
OPA4IN0-	AN22	AN		ADC Channel 22 input.					
	DAC4OUT1		AN	DAC4 voltage output.					
	OPA4IN0-	AN		Operational amplifier 4 inverting input.					
RD3/AN23/C8IN2-	RD3	TTL/ST	CMOS	General purpose I/O.					
	AN23	AN23 AN		ADC Channel 23 input.					
	C8IN2-	AN		Comparator 8 negative input.					
RD4/AN24/C7IN2-	RD4	TTL/ST	CMOS	General purpose I/O.					
	AN24	AN24 AN — ADC Channel 24 input.							
	C7IN2-	AN		Comparator 8 negative input.					
RD5/AN25/C7IN3-/C8IN3-	RD5	TTL/ST	CMOS	General purpose I/O.					
	AN25	AN		ADC Channel 25 input.					
	C7IN3-	AN		Comparator 7 negative input.					
	C8IN3-	AN		Comparator 8 negative input.					
RD6/AN26/C7IN1+	RD6	TTL/ST	CMOS	General purpose I/O.					
	AN26	AN		ADC Channel 26 input.					
	C7IN1+	AN	_	Comparator 7 positive input.					
RD7/AN27/C8IN1+	RD7	TTL/ST	CMOS	General purpose I/O.					
	AN27	AN	—	ADC Channel 27 input.					
	C8IN1+	AN		Comparator 8 positive input.					
RE0/AN5/DAC6REF1+/	RE0	TTL/ST	CMOS	General purpose I/O.					
DAC8REF1+	AN5	AN	—	ADC Channel 5 input.					
	DAC6REF1+	AN		DAC6 positive reference.					
	DAC8REF1+	AN		DAC8 positive reference.					

Legend:AN = Analog input or outputCMOS = CMOS compatible input or outputOD = Open-DrainTTL = TTL compatible inputST = Schmitt Trigger input with CMOS levels I^2C = Schmitt Trigger input with I²C

HP = High Power XTAL = Crystal levels **Note 1:** Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.

2: All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.

Name	Function	Input Type	Output Type	Description					
RE1/AN6/DAC6OUT1/	RE1	TTL/ST	CMOS	General purpose I/O.					
DAC6REF1-/DAC8REF1-	AN6	AN		ADC Channel 6 input.					
	DAC6OUT1	_	AN	DAC6 voltage output.					
	DAC6REF1-	AN		DAC6 negative reference.					
	DAC8REF1-	AN	_	DAC8 negative reference.					
RE2/AN7/DAC8OUT1	RE2	TTL/ST	CMOS	General purpose I/O.					
	AN7	AN	_	ADC Channel 7 input.					
	DAC8OUT1	—	AN	DAC8 voltage output.					
RE3/MCLR/VPP	RE3	TTL/ST	CMOS	General purpose input.					
	MCLR	ST	_	Master clear input.					
Vdd	Vdd								
Vss	Vss	Power		Ground reference.					
OUT ⁽²⁾	C1OUT		CMOS	Comparator 1 output.					
	C2OUT		CMOS	Comparator 2 output.					
	C3OUT		CMOS	Comparator 3 output.					
	C4OUT		CMOS	Comparator 4 output.					
	C5OUT		CMOS	Comparator 5 output.					
	C6OUT		CMOS	Comparator 6 output.					
	C7OUT		CMOS	Comparator 7 output.					
	C8OUT		CMOS	Comparator 8 output.					
	CCP1		CMOS	Compare/PWM1 output.					
	CCP2		CMOS	Compare/PWM2 output.					
	CCP7		CMOS	Compare/PWM7 output.					
	CCP8		CMOS	Compare/PWM8 output.					
	MD1OUT		CMOS	Data signal modulator 1 output.					
	MD2OUT		CMOS	Data signal modulator 2 output.					
	MD3OUT		CMOS	Data signal modulator 3 output.					
	MD4OUT		CMOS	Data signal modulator 4 output.					
	PWM3OUT		CMOS	PWM3 output.					
	PWM4OUT		CMOS	PWM4 output.					
	PWM5OUT		CMOS	PWM5 output.					
	PWM6OUT		CMOS	PWM6 output.					
	PWM9OUT		CMOS	PWM9 output.					
	PWM10OUT		CMOS	PWM10 output.					
	PWM11OUT		CMOS	PWM11 output.					
	PWM12OUT		CMOS	PWM12 output.					
	COG1A		CMOS	Complementary output generator 1 output A.					
	COG1B		CMOS	Complementary output generator 1 output B.					
	COG1C		CMOS	Complementary output generator 1 output C.					
	COG1D		CMOS	Complementary output generator 1 output D.					
	COG2A		CMOS	Complementary output generator 2 output A.					

TABLE 1-3: PIC16(L)F1777/9 PINOUT DESCRIPTION (CONTINUED)

Note 1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.

2: All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.

TADLE 1-3: FIGTO(L)FT////9 FINOUT DESCRIPTION (CONTINUED)	TABLE 1-3:	PIC16(L)F1777/9 PINOUT DESCRIPTION (CONTINUED)
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Name	Function	Input Type	Output Type	Description
OUT ⁽²⁾ (Cont.)	COG2B		CMOS	Complementary output generator 2 output B.
	COG2C		CMOS	Complementary output generator 2 output C.
	COG2D		CMOS	Complementary output generator 2 output D.
	COG3A		CMOS	Complementary output generator 3 output A.
	COG3B		CMOS	Complementary output generator 3 output B.
	COG3C		CMOS	Complementary output generator 3 output C.
	COG3D		CMOS	Complementary output generator 3 output D.
	COG4A		CMOS	Complementary output generator 4 output A.
	COG4B		CMOS	Complementary output generator 4 output B.
	COG4C		CMOS	Complementary output generator 4 output C.
	COG4D		CMOS	Complementary output generator 4 output D.
	SDA ⁽³⁾		OD	I ² C data output.
	SCK		CMOS	SPI clock output.
	SCL ⁽³⁾		OD	I ² C clock output.
	SDO		CMOS	SPI data output.
	TX		CMOS	EUSART asynchronous TX data out.
	СК		CMOS	EUSART synchronous clock out.
	DT ⁽³⁾		CMOS	EUSART synchronous data output.
	CLC1OUT		CMOS	Configurable logic cell 1 output.
	CLC2OUT		CMOS	Configurable logic cell 2 output.
	CLC3OUT		CMOS	Configurable logic cell 3 output.
	CLC4OUT		CMOS	Configurable logic cell 4 output.

 Legend:
 AN = Analog input or output
 CMOS = CMOS compatible input or output
 OD = Open-Drain

 TTL = TTL compatible input
 ST = Schmitt Trigger input with CMOS levels
 I²C = Schmitt Trigger input with I²C

 HP = High Power
 XTAL = Crystal levels

Note 1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.

2: All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.

3: These peripheral functions are bidirectional. The output pin selections must be the same as the input pin selections.

1.2 Peripheral Connection Matrix

Input selection multiplexers on many of the peripherals enable selecting the output of another peripheral such that the signal path is contained entirely within the device. Although the peripheral output can also be routed to a pin, with the PPS selection feature, it is not necessary to do so. Table 1-4 shows all the possible inter-peripheral signal connections. Please refer to corresponding peripheral section to obtain the multiplexer selection codes for the desired connection.

PIC16(L)F1777/8/9

	Peripheral Input																								
Peripheral Output	ADC Trigger	COG Clock	COG Rising/Falling	COG Shutdown	10-bit DAC	5-bit DAC	PRG Analog Input	PRG Rising/Falling	Comparator +	Comparator -	CLC	DSM CH	DSM CL	DSM Mod	Op Amp +	Op Amp -	Op Amp Override	10-bit PWM	16-bit PWM	CCP Capture	CCP Clock	Timer2/4/6/8 Clock	Timer2/4/6/8 Reset	Timer1/3/5 Gate	Timer0 Clock
FVR					٠	٠	•		•	•					٠	٠									
ZCD											•						•					•			
PRG									•						•	•									
10-bit DAC							•		•						•	•								1	
5-bit DAC							•		•						•	•								1	
CCP	•		•					•			•	•	•	•			•						•	1	
Comparator (sync)	•							•			•						•			•			•	•	
Comparator (async)			•	٠										•											
CLC	٠		٠	•							•	٠	•	•			•			•		•	•		
DSM																									
COG																	•								
EUSART TX/CK											٠			•											
EUSART DT											٠			•											
MSSP SCK/SCL											٠			•											
MSSP SDO/SDA											٠			•											
Op Amp							٠																		
10-bit PWM	•		•					•			•	•	•	•			•						•		
16-bit PWM	٠		٠					٠			٠	٠	•	•			•						٠		
Timer0 overflow	٠										٠													٠	
Timer2 = T2PR				٠							•							•			•		•		
Timer4 = T4PR				٠							•							•			•		•		
Timer6 = T6PR				٠							•							•			•		•		
Timer8 = T8PR				٠							•							•			•		•		
Timer2 Postscale	•			٠							•							•			•		•		
Timer4 Postscale	•			٠							•							•			•		•		
Timer6 Postscale	•			٠							•							•			•		•		
Timer8 Postscale	•			٠							•							•			•		•		
Timer1 overflow	•										•							•			•				
Timer3 overflow	•										•							•			•				
Timer5 overflow	•										•							•			•				
SOSC																			•			٠			
Fosc/4		•																				•			
Fosc		•									•	٠	•						•			٠			
HFINTOSC		٠									•	٠	•						•			٠			
LFINTOSC											•								•			٠			
MFINTOSC																						٠			
IOCIF											•									•	•				
PPS Input pin			•	٠				٠				•	•	•						•	٠	•	٠	٠	•

TABLE 1-4: PERIPHERAL CONNECTION MATRIX

2.0 ENHANCED MID-RANGE CPU

This family of devices contain an enhanced mid-range 8-bit CPU core. The CPU has 49 instructions. Interrupt capability includes automatic context saving. The hardware stack is 16 levels deep and has Overflow and Underflow Reset capability. Direct, Indirect, and Relative Addressing modes are available. Two File Select Registers (FSRs) provide the ability to read program and data memory.

- · Automatic Interrupt Context Saving
- 16-level Stack with Overflow and Underflow
- File Select Registers
- Instruction Set

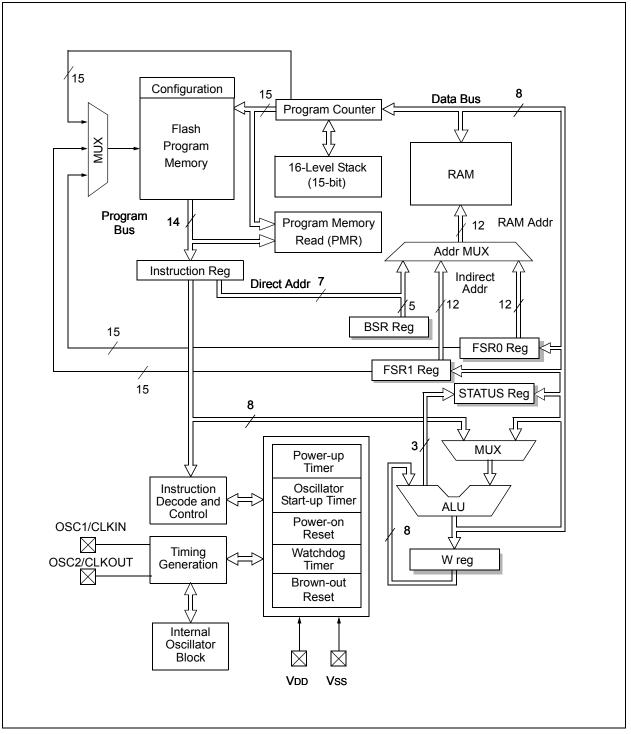


FIGURE 2-1: CORE BLOCK DIAGRAM

2.1 Automatic Interrupt Context Saving

During interrupts, certain registers are automatically saved in shadow registers and restored when returning from the interrupt. This saves stack space and user code. See **Section 7.5 "Automatic Context Saving"** for more information.

2.2 16-Level Stack with Overflow and Underflow

These devices have a hardware stack memory 15 bits wide and 16 words deep. A Stack Overflow or Underflow will set the appropriate bit (STKOVF or STKUNF) in the PCON register, and if enabled, will cause a software Reset. See **Section 3.6 "Stack**" for more details.

2.3 File Select Registers

There are two 16-bit File Select Registers (FSR). FSRs can access all file registers and program memory, which allows one Data Pointer for all memory. When an FSR points to program memory, there is one additional instruction cycle in instructions using INDF to allow the data to be fetched. General purpose memory can now also be addressed linearly, providing the ability to access contiguous data larger than 80 bytes. There are also new instructions to support the FSRs. See **Section 3.7 "Indirect Addressing**" for more details.

2.4 Instruction Set

There are 49 instructions for the enhanced mid-range CPU to support the features of the CPU. See **Section 35.0 "Instruction Set Summary"** for more details.

3.0 MEMORY ORGANIZATION

These devices contain the following types of memory:

- Program Memory
 - Configuration Words
 - Device ID
 - User ID
 - Flash Program Memory
- Data Memory
 - Core Registers
 - Special Function Registers
 - General Purpose RAM
 - Common RAM

Note 1: The method to access Flash memory through the PMCON registers is described in Section 10.0 "Flash Program Memory Control".

The following features are associated with access and control of program memory and data memory:

- PCL and PCLATH
- Stack
- Indirect Addressing

TABLE 3-1: DEVICE SIZES AND ADDRESSES

3.1 **Program Memory Organization**

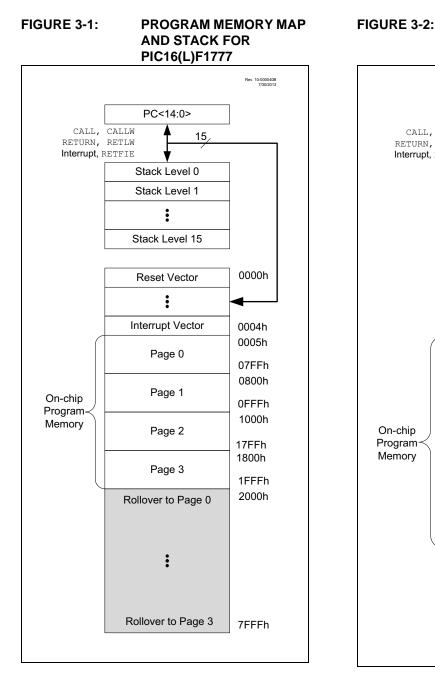
The enhanced mid-range core has a 15-bit program counter capable of addressing a 32K x 14 program memory space. Table 3-1 shows the memory sizes implemented for the PIC16(L)F1777/8/9 family. Accessing a location above these boundaries will cause a wrap-around within the implemented memory space. The Reset vector is at 0000h and the interrupt vector is at 0004h (see Figures 3-1 and 3-2).

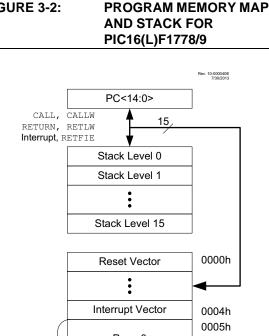
3.2 High-Endurance Flash

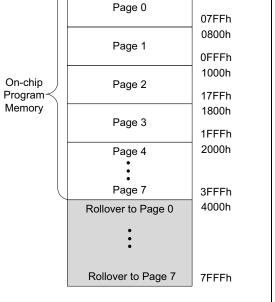
This device has a 128-byte section of high-endurance program Flash memory (PFM) in lieu of data EEPROM. This area is especially well suited for nonvolatile data storage that is expected to be updated frequently over the life of the end product. See **Section 10.2** "**Flash Program Memory Overview**" for more information on writing data to PFM. See **Section 3.2.1.2** "**Indirect Read with FSR**" for more information about using the FSR registers to read byte data stored in PFM.

Device	Program Memory Space (Words)	Last Program Memory Address	High-Endurance Flash Memory Address Range ⁽¹⁾
PIC16(L)F1777	8,192	1FFFh	1F80h-1FFFh
PIC16(L)F1778/9	16,384	3FFFh	3F80h-3FFFh

Note 1: High-endurance Flash applies to the low byte of each address in the range.







3.2.1 READING PROGRAM MEMORY AS DATA

There are two methods of accessing constants in program memory. The first method is to use tables of RETLW instructions. The second method is to set an FSR to point to the program memory.

3.2.1.1 RETLW Instruction

The RETLW instruction can be used to provide access to tables of constants. The recommended way to create such a table is shown in Example 3-1.

EXAMPLE 3-1: RETLW INSTRUCTION

constants	
BRW	;Add Index in W to
	;program counter to
	;select data
RETLW DATA0	;Index0 data
RETLW DATA1	;Index1 data
RETLW DATA2	
RETLW DATA3	
my_function	
; LOTS OF CODE	
MOVLW DATA_IN	DEX
call constants	
; THE CONSTANT IS	IN W
L	

The BRW instruction makes this type of table very simple to implement. If your code must remain portable with previous generations of microcontrollers, then the BRW instruction is not available so the older table read method must be used.

3.2.1.2 Indirect Read with FSR

The program memory can be accessed as data by setting bit 7 of the FSRxH register and reading the matching INDFx register. The MOVIW instruction will place the lower eight bits of the addressed word in the W register. Writes to the program memory cannot be performed via the INDF registers. Instructions that access the program memory via the FSR require one extra instruction cycle to complete. Example 3-2 demonstrates accessing the program memory via an FSR.

The high directive will set bit<7> if a label points to a location in program memory.

EXAMPLE 3-2: ACCESSING PROGRAM MEMORY VIA FSR

constants	
DW DATA0 ;First constant	t
DW DATA1 ;Second constan	nt
DW DATA2	
DW DATA3	
my_function	
; LOTS OF CODE	
MOVLW DATA_INDEX	
ADDLW LOW constants	
MOVWF FSR1L	
MOVLW HIGH constants;MSb sets	
automatically	
MOVWF FSR1H	
BTFSC STATUS, C ;carry from ADI	DLW?
INCF FSR1H, f ;yes	
MOVIW 0[FSR1]	
;THE PROGRAM MEMORY IS IN W	

3.3 Data Memory Organization

The data memory is partitioned in 32 memory banks with 128 bytes in a bank. Each bank consists of (Figure 3-3):

- 12 core registers
- 20 Special Function Registers (SFR)
- · Up to 80 bytes of General Purpose RAM (GPR)
- · 16 bytes of common RAM

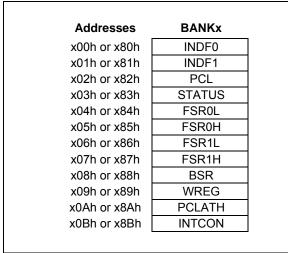
The active bank is selected by writing the bank number into the Bank Select Register (BSR). Unimplemented memory will read as '0'. All data memory can be accessed either directly (via instructions that use the file registers) or indirectly via the two File Select Registers (FSR). See Section 3.7 "Indirect Addressing" for more information.

Data memory uses a 12-bit address. The upper five bits of the address define the Bank address and the lower seven bits select the registers/RAM in that bank.

3.3.1 CORE REGISTERS

The core registers contain the registers that directly affect the basic operation. The core registers occupy the first 12 addresses of every data memory bank (addresses x00h/x08h through x0Bh/x8Bh). These registers are listed below in Table 3-2. For detailed information, see Table 3-17.

TABLE 3-2: CORE REGISTERS



3.3.1.1 STATUS Register

The STATUS register, shown in Register 3-1, contains:

- the arithmetic status of the ALU
- · the Reset status

The STATUS register can be the destination for any instruction, like any other register. If the STATUS register is the destination for an instruction that affects the Z, DC or C bits, then the write to these three bits is disabled. These bits are set or cleared according to the device logic. Furthermore, the TO and PD bits are not writable. Therefore, the result of an instruction with the STATUS register as destination may be different than intended.

For example, CLRF STATUS will clear the upper three bits and set the Z bit. This leaves the STATUS register as '000u uluu' (where u = unchanged).

It is recommended, therefore, that only BCF, BSF, SWAPF and MOVWF instructions are used to alter the STATUS register, because these instructions do not affect any Status bits. For other instructions not affecting any Status bits (Refer to Section 35.0 "Instruction Set Summary").

Note: The C and DC bits operate as Borrow and Digit Borrow out bits, respectively, in subtraction.

PIC16(L)F1777/8/9

3.4 Register Definitions: Status

REGISTER 3-1: STATUS: STATUS REGISTER

U-0	U-0	U-0	R-1/q	R-1/q	R/W-0/u	R/W-0/u	R/W-0/u
_	_	_	TO	PD	Z	DC ⁽¹⁾	C ⁽¹⁾
bit 7	•						bit 0
Legend:							
R = Readable	bit	W = Writable	bit	U = Unimpler	mented bit, read	as '0'	
u = Bit is unch	anged	x = Bit is unkr	nown	-n/n = Value	at POR and BOI	R/Value at all o	ther Resets
'1' = Bit is set		'0' = Bit is clea	ared	q = Value de	pends on conditi	ion	
bit 7-5	Unimplemen	ted: Read as '	כ'				
bit 4	TO: Time-Out	t bit					
		er-up, CLRWDT		SLEEP instruc	tion		
	0 = A WDT T	ime-out occurre	ed				
bit 3	PD: Power-D	own bit					
		er-up or by the					
	•	tion of the SLEI	EP Instruction				
bit 2	Z: Zero bit			<i></i> .			
		t of an arithmet t of an arithmet			ero		
bit 1			÷ .		SUBWF instruction	one)(1)	
DIL I	•	ut from the 4th				5113)**	
		out from the 4th			curreu		
bit 0	-	ow bit (ADDWF,			ructions) ⁽¹⁾		
	•	ut from the Mos			,		
		out from the Mo	U U				
Note 1: For	Borrow the po	larity is reverse	d A subtract	ion is executed	by adding the t	wo's compleme	ent of the

Note 1: For Borrow, the polarity is reversed. A subtraction is executed by adding the two's complement of the second operand.

3.4.1 SPECIAL FUNCTION REGISTER

The Special Function Registers (SFR) are registers used by the application to control the desired operation of peripheral functions in the device. The SFR occupies the 20 bytes after the core registers of every data memory bank (addresses x0Ch/x8Ch through x1Fh/x9Fh). The registers associated with the operation of each peripheral are described in the corresponding peripheral chapters of this data sheet.

3.4.2 GENERAL PURPOSE RAM

There are up to 80 bytes of General Purpose Registers (GPR) in each data memory bank. The GPR occupies the space immediately after the SFR of selected data memory banks. The number of banks selected depends on the total amount of GPR space available in the device.

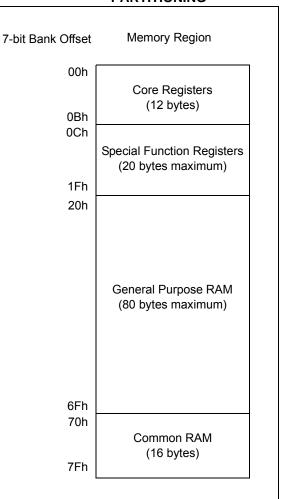
3.4.2.1 Linear Access to GPR

The general purpose RAM can be accessed in a non-banked method via the FSRs. This can simplify access to large memory structures. See **Section 3.7.2** "Linear Data Memory" for more information.

3.4.3 COMMON RAM

There are 16 bytes of common RAM accessible from all banks.

FIGURE 3-3: BANKED MEMORY PARTITIONING



3.4.4 DEVICE MEMORY MAPS

The memory maps for the device family are as shown in Tables 3-3 through 3-16.

TABLE 3-3: PIC16(L)F1778 MEMORY MAP (BANKS 0-7)

	BANK 0		BANK 1		BANK 2		BANK 3		BANK 4		BANK 5		BANK 6		BANK 7
000h		080h		100h		180h		200h		280h		300h		380h	
	Core Registers (Table 3-2)		Core Registers (Table 3-2)		Core Registers (Table 3-2)										
00Bh		08Bh		10Bh		18Bh		20Bh		28Bh		30Bh		38Bh	
00Ch	PORTA	08Ch	TRISA	10Ch	LATA	18Ch	ANSELA	20Ch	WPUA	28Ch	ODCONA	30Ch	SLRCONA	38Ch	INLVLA
00Dh	PORTB	08Dh	TRISB	10Dh	LATB	18Dh	ANSELB	20Dh	WPUB	28Dh	ODCONB	30Dh	SLRCONB	38Dh	INLVLB
00Eh	PORTC	08Eh	TRISC	10Eh	LATC	18Eh	ANSELC	20Eh	WPUC	28Eh	ODCONC	30Eh	SLRCONC	38Eh	INLVLC
00Fh	_	08Fh	_	10Fh	_	18Fh	_	20Fh	_	28Fh	_	30Fh	_	38Fh	
010h	PORTE	090h	TRISE	110h	_	190h	—	210h	WPUE	290h	—	310h	—	390h	INLVLE
011h	PIR1	091h	PIE1	111h	CMOUT	191h	PMADRL	211h	SSP1BUF	291h	CCPR1L	311h	—	391h	IOCAP
012h	PIR2	092h	PIE2	112h	CM1CON0	192h	PMADRH	212h	SSP1ADD	292h	CCPR1H	312h	—	392h	IOCAN
013h	PIR3	093h	PIE3	113h	CM1CON1	193h	PMDATL	213h	SSP1MSK	293h	CCP1CON	313h	—	393h	IOCAF
014h	PIR4	094h	PIE4	114h	CM1NSEL	194h	PMDATH	214h	SSP1STAT	294h	CCP1CAP	314h	-	394h	IOCBP
015h	PIR5	095h	PIE5	115h	CM1PSEL	195h	PMCON1	215h	SSP1CON1	295h	CCPR2L	315h	MD1CON0	395h	IOCBN
016h	PIR6	096h	PIE6	116h	CM2CON0	196h	PMCON2	216h	SSP1CON2	296h	CCPR2H	316h	MD1CON1	396h	IOCBF
017h	TMR0	097h	OPTION_REG	117h	CM2CON1	197h	VREGCON ⁽¹⁾	217h	SSP1CON3	297h	CCP2CON	317h	MD1SRC	397h	IOCCP
018h	TMR1L	098h	PCON	118h	CM2NSEL	198h	—	218h	—	298h	CCP2CAP	318h	MD1CARL	398h	IOCCN
019h	TMR1H	099h	WDTCON	119h	CM2PSEL	199h	RC1REG	219h	_	299h	CCPR7L	319h	MD1CARH	399h	IOCCF
01Ah	T1CON	09Ah	OSCTUNE	11Ah	CM3CON0	19Ah	TX1REG	21Ah	—	29Ah	CCPR7H	31Ah	—	39Ah	_
01Bh	T1GCON	09Bh	OSCCON	11Bh	CM3CON1	19Bh	SP1BRGL	21Bh	MD3CON0	29Bh	CCP7CON	31Bh	MD2CON0	39Bh	
01Ch	TMR3L	09Ch	OSCSTAT	11Ch	CM3NSEL	19Ch	SP1BRGH	21Ch	MD3CON1	29Ch	CCP7CAP	31Ch	MD2CON1	39Ch	
01Dh	TMR3H	09Dh	BORCON	11Dh	CM3PSEL	19Dh	RC1STA	21Dh	MD3SRC	29Dh	—	31Dh	MD2SRC	39Dh	IOCEP
01Eh	T3CON	09Eh	FVRCON	11Eh	—	19Eh	TX1STA	21Eh	MD3CARL	29Eh	CCPTMRS1	31Eh	MD2CARL	39Eh	IOCEN
01Fh	T3GCON	09Fh	ZCD1CON	11Fh	—	19Fh	BAUD1CON	21Fh	MD3CARH	29Fh	CCPTMRS2	31Fh	MD2CARH	39Fh	IOCEF
020h		0A0h		120h		1A0h		220h		2A0h		320h		3A0h	
	General Purpose Register	32Fh 330h	General Purpose Register		General Purpose Register										
	80 Bytes		80 Bytes		80 Bytes										
06Fh		0EFh		16Fh		1EFh		26Fh		2EFh		36Fh		3EFh	
070h		0F0h		170h		1F0h		270h		2F0h		370h		3F0h	
	Common RAM 70h – 7Fh		Accesses 70h – 7Fh		Accesses 70h – 7Fh		Accesses 70h – 7Fh		Accesses 70h – 7Fh		Accesses 70h – 7Fh		Accesses 70h – 7Fh		Accesses 70h – 7Fh
07Fh		0FFh		17Fh		1FFh		27Fh		2FFh		37Fh		3FFh	

Legend: = Unimplemented data memory locations, read as '0'.

Note 1: Unimplemented on PIC16LF1778.

TABLE 3-4:PIC16(L)F1777/9 MEMORY MAP (BANKS 0-7)

	BANK 0		BANK 1		BANK 2		BANK 3		BANK 4		BANK 5		BANK 6		BANK 7
000h	Core Registers (Table 3-2)	080h	Core Registers (Table 3-2)	100h	Core Registers (Table 3-2)	180h	Core Registers (Table 3-2)	200h	Core Registers (Table 3-2)	280h	Core Registers (Table 3-2)	300h	Core Registers (Table 3-2)	380h	Core Registers (Table 3-2)
00Bh		08Bh		10Bh		18Bh		20Bh		28Bh		30Bh		38Bh	
00Ch	PORTA	08Ch	TRISA	10Ch	LATA	18Ch	ANSELA	20Ch	WPUA	28Ch	ODCONA	30Ch	SLRCONA	38Ch	INLVLA
00Dh	PORTB	08Dh	TRISB	10Dh	LATB	18Dh	ANSELB	20Dh	WPUB	28Dh	ODCONB	30Dh	SLRCONB	38Dh	INLVLB
00Eh	PORTC	08Eh	TRISC	10Eh	LATC	18Eh	ANSELC	20Eh	WPUC	28Eh	ODCONC	30Eh	SLRCONC	38Eh	INLVLC
00Fh	PORTD	08Fh	TRISD	10Fh	LATD	18Fh	ANSELD	20Fh	WPUD	28Fh	ODCOND	30Fh	SLRCOND	38Fh	INLVLD
010h	PORTE	090h	TRISE	110h	LATE	190h	ANSELE	210h	WPUE	290h	ODCONE	310h	SLRCONE	390h	INLVLE
011h	PIR1	091h	PIE1	111h	CMOUT	191h	PMADRL	211h	SSP1BUF	291h	CCPR1L	311h	CCPR8L	391h	IOCAP
012h	PIR2	092h	PIE2	112h	CM1CON0	192h	PMADRH	212h	SSP1ADD	292h	CCPR1H	312h	CCPR8H	392h	IOCAN
013h	PIR3	093h	PIE3	113h	CM1CON1	193h	PMDATL	213h	SSP1MSK	293h	CCP1CON	313h	CCP8CON	393h	IOCAF
014h	PIR4	094h	PIE4	114h	CM1NSEL	194h	PMDATH	214h	SSP1STAT	294h	CCP1CAP	314h	CCP8CAP	394h	IOCBP
015h	PIR5	095h	PIE5	115h	CM1PSEL	195h	PMCON1	215h	SSP1CON1	295h	CCPR2L	315h	MD1CON0	395h	IOCBN
016h	PIR6	096h	PIE6	116h	CM2CON0	196h	PMCON2	216h	SSP1CON2	296h	CCPR2H	316h	MD1CON1	396h	IOCBF
017h	TMR0	097h	OPTION_REG	117h	CM2CON1	197h	VREGCON ⁽¹⁾	217h	SSP1CON3	297h	CCP2CON	317h	MD1SRC	397h	IOCCP
018h	TMR1L	098h	PCON	118h	CM2NSEL	198h	_	218h	—	298h	CCP2CAP	318h	MD1CARL	398h	IOCCN
019h	TMR1H	099h	WDTCON	119h	CM2PSEL	199h	RC1REG	219h	—	299h	CCPR7L	319h	MD1CARH	399h	IOCCF
01Ah	T1CON	09Ah	OSCTUNE	11Ah	CM3CON0	19Ah	TX1REG	21Ah	—	29Ah	CCPR7H	31Ah	—	39Ah	_
01Bh	T1GCON	09Bh	OSCCON	11Bh	CM3CON1	19Bh	SP1BRGL	21Bh	MD3CON0	29Bh	CCP7CON	31Bh	MD2CON0	39Bh	_
01Ch	TMR3L	09Ch	OSCSTAT	11Ch	CM3NSEL	19Ch	SP1BRGH	21Ch	MD3CON1	29Ch	CCP7CAP	31Ch	MD2CON1	39Ch	
01Dh	TMR3H	09Dh	BORCON	11Dh	CM3PSEL	19Dh	RC1STA	21Dh	MD3SRC	29Dh	—	31Dh	MD2SRC	39Dh	IOCEP
01Eh	T3CON	09Eh	FVRCON	11Eh		19Eh	TX1STA	21Eh	MD3CARL	29Eh	CCPTMRS1	31Eh	MD2CARL	39Eh	IOCEN
01Fh	T3GCON	09Fh	ZCD1CON	11Fh	—	19Fh	BAUD1CON	21Fh	MD3CARH	29Fh	CCPTMRS2	31Fh	MD2CARH	39Fh	IOCEF
020h		0A0h		120h		1A0h		220h		2A0h		320h		3A0h	
	General Purpose Register 80 Bytes														
06Fh		0EFh		16Fh		1EFh		26Fh		2EFh		36Fh		3EFh	
070h 07Fh	Common RAM 70h – 7Fh	0F0h 0FFh	Accesses 70h – 7Fh	170h 17Fh	Accesses 70h – 7Fh	1F0h 1FFh	Accesses 70h – 7Fh	270h 27Fh	Accesses 70h – 7Fh	2F0h 2FFh	Accesses 70h – 7Fh	370h 37Fh	Accesses 70h – 7Fh	3F0h 3FFh	Accesses 70h – 7Fh
5/111						[<i></i>			

Legend: = Unimplemented data memory locations, read as '0'.

Note 1: Unimplemented on PIC16LF1777/9.

TABLE 3-5:PIC16(L)F1778 MEMORY MAP, BANK 8-15

	BANK 8		BANK 9		BANK 10		BANK 11		BANK 12		BANK 13		BANK 14		BANK 15
400h		480h		500h		580h		600h		680h		700h		780h	
	Core Registers														
	(Table 3-2)														
40Bh		48Bh		50Bh		58Bh		60Bh		68Bh		70Bh		78Bh	
40Ch	—	48Ch	—	50Ch	—	58Ch	—	60Ch	—	68Ch	—	70Ch	—	78Ch	_
40Dh	HIDRVB	48Dh	_	50Dh	—	58Dh	DACLD	60Dh	—	68Dh	COG1PHR	70Dh	COG2PHR	78Dh	_
40Eh	—	48Eh	ADRESL	50Eh	—	58Eh	DAC1CON0	60Eh	—	68Eh	COG1PHF	70Eh	COG2PHF	78Eh	PRG1RTSS
40Fh	TMR5L	48Fh	ADRESH	50Fh	OPA1NCHS	58Fh	DAC1REFL	60Fh	—	68Fh	COG1BLKR	70Fh	COG2BLKR	78Fh	PRG1FTSS
410h	TMR5H	490h	ADCON0	510h	OPA1PCHS	590h	DAC1REFH	610h	—	690h	COG1BLKF	710h	COG2BLKF	790h	PRG1INS
411h	T5CON	491h	ADCON1	511h	OPA1CON	591h	DAC2CON0	611h	—	691h	COG1DBR	711h	COG2DBR	791h	PRG1CON0
412h	T5GCON	492h	ADCON2	512h	OPA10RS	592h	DAC2REFL	612h	—	692h	COG1DBF	712h	COG2DBF	792h	PRG1CON1
413h	T4TMR	493h	T2TMR	513h	OPA2NCHS	593h	DAC2REFH	613h	—	693h	COG1CON0	713h	COG2CON0	793h	PRG1CON2
414h	T4PR	494h	T2PR	514h	OPA2PCHS	594h	DAC3CON0	614h	PWM3DCL	694h	COG1CON1	714h	COG2CON1	794h	PRG2RTSS
415h	T4CON	495h	T2CON	515h	OPA2CON	595h	DAC3REF	615h	PWM3DCH	695h	COG1RIS0	715h	COG2RIS0	795h	PRG2FTSS
416h	T4HLT	496h	T2HLT	516h	OPA2ORS	596h	DAC4CON0	616h	PWM3CON	696h	COG1RIS1	716h	COG2RIS1	796h	PRG2INS
417h	T4CLKCON	497h	T2CLKCON	517h	OPA3NCHS	597h	DAC4REF	617h	PWM4DCL	697h	COG1RSIM0	717h	COG2RSIM0	797h	PRG2CON0
418h	T4RST	498h	T2RST	518h	OPA3PCHS	598h	DAC5CON0	618h	PWM4DCH	698h	COG1RSIM1	718h	COG2RSIM1	798h	PRG2CON1
419h	_	499h	_	519h	OPA3CON	599h	DAC5REFL	619h	PWM4CON	699h	COG1FIS0	719h	COG2FIS0	799h	PRG2CON2
41Ah	T6TMR	49Ah	T8TMR	51Ah	OPA3ORS	59Ah	DAC5REFH	61Ah	PWM9DCL	69Ah	COG1FIS1	71Ah	COG2FIS1	79Ah	PRG3RTSS
41Bh	T6PR	49Bh	T8PR	51Bh	—	59Bh	—	61Bh	PWM9DCH	69Bh	COG1FSIM0	71Bh	COG2FSIM0	79Bh	PRG3FTSS
41Ch	T6CON	49Ch	T8CON	51Ch	_	59Ch	_	61Ch	PWM9CON	69Ch	COG1FSIM1	71Ch	COG2FSIM1	79Ch	PRG3INS
41Dh	T6HLT	49Dh	T8HLT	51Dh	—	59Dh	—	61Dh	—	69Dh	COG1ASD0	71Dh	COG2ASD0	79Dh	PRG3CON0
41Eh	T6CLKCON	49Eh	T8CLKCON	51Eh	—	59Eh	DAC7CON0	61Eh	—	69Eh	COG1ASD1	71Eh	COG2ASD1	79Eh	PRG3CON1
41Fh	T6RST	49Fh	T8RST	51Fh	_	59Fh	DA73REF	61Fh	—	69Fh	COG1STR	71Fh	COG2STR	79Fh	PRG3CON2
420h		4A0h		520h		5A0h		620h		6A0h		720h		7A0h	
	General														
	Purpose														
	Register														
105	80 Bytes		80 Bytes		80 Bytes		80 Bytes	005	80 Bytes		80 Bytes		80 Bytes		80 Bytes
46Fh		4EFh		56Fh		5EFh		66Fh		6EFh		76Fh		7EFh	
470h		4F0h		570h		5F0h		670h		6F0h		770h		7F0h	
	Accesses														
	70h – 7Fh														
47Fh		4FFh		57Fh		5FFh		67Fh		6FFh		77Fh		7FFh	

TABLE 3-6: PIC16(L)F1777 MEMORY MAP, BANK 8-15

	BANK 8		BANK 9		BANK 10		BANK 11		BANK 12		BANK 13		BANK 14		BANK 15
400h		480h		500h		580h		600h		680h		700h		780h	
	Core Registers		Core Registers		Core Registers		Core Registers		Core Registers		Core Registers		Core Registers		Core Registers
	(Table 3-2)		(Table 3-2)		(Table 3-2)		(Table 3-2)		(Table 3-2)		(Table 3-2)		(Table 3-2)		(Table 3-2)
40Bh		48Bh		50Bh		58Bh		60Bh		68Bh		70Bh		78Bh	
40Ch	—	48Ch	—	50Ch	—	58Ch	—	60Ch	DAC8CON0	68Ch	—	70Ch	_	78Ch	—
40Dh	HIDRVB	48Dh	—	50Dh	—	58Dh	DACLD	60Dh	DAC8REFL	68Dh	COG1PHR	70Dh	COG2PHR	78Dh	—
40Eh	—	48Eh	ADRESL	50Eh	—	58Eh	DAC1CON0	60Eh	PRG4RTSS	68Eh	COG1PHF	70Eh	COG2PHF	78Eh	PRG1RTSS
40Fh	TMR5L	48Fh	ADRESH	50Fh	OPA1NCHS	58Fh	DAC1REFL	60Fh	PRG4FTSS	68Fh	COG1BLKR	70Fh	COG2BLKR	78Fh	PRG1FTSS
410h	TMR5H	490h	ADCON0	510h	OPA1PCHS	590h	DAC1REFH	610h	PRG4INS	690h	COG1BLKF	710h	COG2BLKF	790h	PRG1INS
411h	T5CON	491h	ADCON1	511h	OPA1CON	591h	DAC2CON0	611h	PRG4CON0	691h	COG1DBR	711h	COG2DBR	791h	PRG1CON0
412h	T5GCON	492h	ADCON2	512h	OPA10RS	592h	DAC2REFL	612h	PRG4CON1	692h	COG1DBF	712h	COG2DBF	792h	PRG1CON1
413h	T4TMR	493h	T2TMR	513h	OPA2NCHS	593h	DAC2REFH	613h	PRG4CON2	693h	COG1CON0	713h	COG2CON0	793h	PRG1CON2
414h	T4PR	494h	T2PR	514h	OPA2PCHS	594h	DAC3CON0	614h	PWM3DCL	694h	COG1CON1	714h	COG2CON1	794h	PRG2RTSS
415h	T4CON	495h	T2CON	515h	OPA2CON	595h	DAC3REF	615h	PWM3DCH	695h	COG1RIS0	715h	COG2RIS0	795h	PRG2FTSS
416h	T4HLT	496h	T2HLT	516h	OPA2ORS	596h	DAC4CON0	616h	PWM3CON	696h	COG1RIS1	716h	COG2RIS1	796h	PRG2INS
417h	T4CLKCON	497h	T2CLKCON	517h	OPA3NCHS	597h	DAC4REF	617h	PWM4DCL	697h	COG1RSIM0	717h	COG2RSIM0	797h	PRG2CON0
418h	T4RST	498h	T2RST	518h	OPA3PCHS	598h	DAC5CON0	618h	PWM4DCH	698h	COG1RSIM1	718h	COG2RSIM1	798h	PRG2CON1
419h	—	499h	—	519h	OPA3CON	599h	DAC5REFL	619h	PWM4CON	699h	COG1FIS0	719h	COG2FIS0	799h	PRG2CON2
41Ah	T6TMR	49Ah	T8TMR	51Ah	OPA3ORS	59Ah	DAC5REFH	61Ah	PWM9DCL	69Ah	COG1FIS1	71Ah	COG2FIS1	79Ah	PRG3RTSS
41Bh	T6PR	49Bh	T8PR	51Bh	OPA4NCHS	59Bh	DAC6CON0	61Bh	PWM9DCH	69Bh	COG1FSIM0	71Bh	COG2FSIM0	79Bh	PRG3FTSS
41Ch	T6CON	49Ch	T8CON	51Ch	OPA4PCHS	59Ch	DAC6REFL	61Ch	PWM9CON	69Ch	COG1FSIM1	71Ch	COG2FSIM1	79Ch	PRG3INS
41Dh	T6HLT	49Dh	T8HLT	51Dh	OPA4CON	59Dh	DAC6REFH	61Dh	PWM10DCL	69Dh	COG1ASD0	71Dh	COG2ASD0	79Dh	PRG3CON0
41Eh	T6CLKCON	49Eh	T8CLKCON	51Eh	OPA4ORS	59Eh	DAC7CON0	61Eh	PWM10DCH	69Eh	COG1ASD1	71Eh	COG2ASD1	79Eh	PRG3CON1
41Fh	T6RST	49Fh 4A0h	T8RST	51Fh	—	59Fh 5A0h	DA73REF	61Fh 620h	PWM10CON	69Fh 6A0h	COG1STR	71Fh 720h	COG2STR	79Fh	PRG3CON2
420h	a	4A011	a .	520h	. .	SAUN	a .	02011	General Purpose	6A01		72011		7A0h	
	General		General		General		General	64Fh	Register 48 Bytes		l la inca la un e a te el		l la inca la un e a te el		L la inca la manda d
	Purpose Register		Purpose Register		Purpose Register		Purpose Register	650h			Unimplemented Read as '0'		Unimplemented Read as '0'		Unimplemented Read as '0'
	80 Bytes		80 Bytes		80 Bytes		80 Bytes	05011	Unimplemented		Reau as 0		Reau as 0		Redu as 0
46Fh	00 Dyies	4EFh	00 Dyies	56Fh	00 Dyies	5EFh	00 Dytes	66Fh	Read as '0'	6EFh		76Fh		7EFh	
470h		4F0h		570h		5F0h		670h		6F0h		770h		7F0h	
-1011	A	11 011	A	57.011	A	51 011	A	07011	A	51 011	A		A	11 011	A
	Accesses 70h – 7Fh		Accesses 70h – 7Fh		Accesses 70h – 7Fh		Accesses 70h – 7Fh		Accesses 70h – 7Fh		Accesses 70h – 7Fh		Accesses 70h – 7Fh		Accesses 70h – 7Fh
							1011-1711		1011 - 1711		1011 - 1711				7011 - 7611
47Fh		4FFh		57Fh		5FFh		67Fh		6FFh		77Fh		7FFh	

TABLE 3-7: PIC16(L)F1779 MEMORY MAP, BANK 8-15

	BANK 8		BANK 9		BANK 10		BANK 11		BANK 12		BANK 13		BANK 14		BANK 15
400h		480h		500h		580h		600h		680h		700h		780h	
	Core Registers		Core Registers		Core Registers		Core Registers		Core Registers		Core Registers		Core Registers		Core Register
	(Table 3-2)		(Table 3-2)		(Table 3-2)		(Table 3-2)		(Table 3-2)		(Table 3-2)		(Table 3-2)		(Table 3-2)
40Bh		48Bh		50Bh		58Bh		60Bh		68Bh		70Bh		78Bh	
40Ch	—	48Ch	—	50Ch	—	58Ch	—	60Ch	DAC8CON0	68Ch	—	70Ch	—	78Ch	
10Dh	HIDRVB	48Dh	_	50Dh	—	58Dh	DACLD	60Dh	DAC8REFL	68Dh	COG1PHR	70Dh	COG2PHR	78Dh	—
40Eh	—	48Eh	ADRESL	50Eh	—	58Eh	DAC1CON0	60Eh	PRG4RTSS	68Eh	COG1PHF	70Eh	COG2PHF	78Eh	PRG1RTSS
40Fh	TMR5L	48Fh	ADRESH	50Fh	OPA1NCHS	58Fh	DAC1REFL	60Fh	PRG4FTSS	68Fh	COG1BLKR	70Fh	COG2BLKR	78Fh	PRG1FTSS
410h	TMR5H	490h	ADCON0	510h	OPA1PCHS	590h	DAC1REFH	610h	PRG4INS	690h	COG1BLKF	710h	COG2BLKF	790h	PRG1INS
411h	T5CON	491h	ADCON1	511h	OPA1CON	591h	DAC2CON0	611h	PRG4CON0	691h	COG1DBR	711h	COG2DBR	791h	PRG1CON0
412h	T5GCON	492h	ADCON2	512h	OPA10RS	592h	DAC2REFL	612h	PRG4CON1	692h	COG1DBF	712h	COG2DBF	792h	PRG1CON1
413h	T4TMR	493h	T2TMR	513h	OPA2NCHS	593h	DAC2REFH	613h	PRG4CON2	693h	COG1CON0	713h	COG2CON0	793h	PRG1CON2
414h	T4PR	494h	T2PR	514h	OPA2PCHS	594h	DAC3CON0	614h	PWM3DCL	694h	COG1CON1	714h	COG2CON1	794h	PRG2RTSS
415h	T4CON	495h	T2CON	515h	OPA2CON	595h	DAC3REF	615h	PWM3DCH	695h	COG1RIS0	715h	COG2RIS0	795h	PRG2FTSS
416h	T4HLT	496h	T2HLT	516h	OPA2ORS	596h	DAC4CON0	616h	PWM3CON	696h	COG1RIS1	716h	COG2RIS1	796h	PRG2INS
417h	T4CLKCON	497h	T2CLKCON	517h	OPA3NCHS	597h	DAC4REF	617h	PWM4DCL	697h	COG1RSIM0	717h	COG2RSIM0	797h	PRG2CON
418h	T4RST	498h	T2RST	518h	OPA3PCHS	598h	DAC5CON0	618h	PWM4DCH	698h	COG1RSIM1	718h	COG2RSIM1	798h	PRG2CON1
419h	_	499h	_	519h	OPA3CON	599h	DAC5REFL	619h	PWM4CON	699h	COG1FIS0	719h	COG2FIS0	799h	PRG2CON2
41Ah	T6TMR	49Ah	T8TMR	51Ah	OPA3ORS	59Ah	DAC5REFH	61Ah	PWM9DCL	69Ah	COG1FIS1	71Ah	COG2FIS1	79Ah	PRG3RTSS
41Bh	T6PR	49Bh	T8PR	51Bh	OPA4NCHS	59Bh	DAC6CON0	61Bh	PWM9DCH	69Bh	COG1FSIM0	71Bh	COG2FSIM0	79Bh	PRG3FTSS
41Ch	T6CON	49Ch	T8CON	51Ch	OPA4PCHS	59Ch	DAC6REFL	61Ch	PWM9CON	69Ch	COG1FSIM1	71Ch	COG2FSIM1	79Ch	PRG3INS
41Dh	T6HLT	49Dh	T8HLT	51Dh	OPA4CON	59Dh	DAC6REFH	61Dh	PWM10DCL	69Dh	COG1ASD0	71Dh	COG2ASD0	79Dh	PRG3CON0
41Eh	T6CLKCON	49Eh	T8CLKCON	51Eh	OPA4ORS	59Eh	DAC7CON0	61Eh	PWM10DCH	69Eh	COG1ASD1	71Eh	COG2ASD1	79Eh	PRG3CON1
41Fh	T6RST	49Fh	T8RST	51Fh	_	59Fh	DA73REF	61Fh	PWM10CON	69Fh	COG1STR	71Fh	COG2STR	79Fh	PRG3CON2
420h		4A0h		520h		5A0h		620h		6A0h		720h		7A0h	
	General		General		General		General		General		General		General		General
	Purpose		Purpose		Purpose		Purpose		Purpose		Purpose		Purpose		Purpose
	Register		Register		Register		Register		Register		Register		Register		Register
	80 Bytes		80 Bytes		80 Bytes		80 Bytes		80 Bytes		80 Bytes		80 Bytes		80 Bytes
16Fh		4EFh		56Fh		5EFh		66Fh		6EFh		76Fh		7EFh	
470h		4F0h		570h		5F0h		670h		6F0h		770h		7F0h	
	Accesses		Accesses		Accesses		Accesses		Accesses		Accesses		Accesses		Accesses
	70h – 7Fh		70h – 7Fh		70h – 7Fh		70h – 7Fh		70h – 7Fh		70h – 7Fh		70h – 7Fh		70h – 7Fh
47Fh		4FFh		57Fh		5FFh		67Fh		6FFh		77Fh		7FFh	

TABLE 3-8: PIC16(L)F1778 MEMORY MAP, BANK 16-23

	BANK 16		BANK 17		BANK 18		BANK 19		BANK 20		BANK 21		BANK 22		BANK 23
800h	Core Registers (Table 3-2)	880h	Core Registers (Table 3-2)	900h	Core Registers (Table 3-2)	980h	Core Registers (Table 3-2)	A00h	Core Registers (Table 3-2)	A80h	Core Registers (Table 3-2)	B00h	Core Registers (Table 3-2)	B80h	Core Registers (Table 3-2)
80Bh		88Bh		90Bh		98Bh		A0Bh		A8Bh		B0Bh		B8Bh	
80Ch	_	88Ch		90Ch	CM4CON0	98Ch		A0Ch		A8Ch		B0Ch		B8Ch	
80Dh	COG3PHR			90Dh	CM4CON1										
80Eh	COG3PHF			90Eh	CM4NSEL										
80Fh	COG3BLKR			90Fh	CM4PSEL										
810h	COG3BLKF			910h	CM5CON0										
811h	COG3DBR			911h	CM5CON1										
812h	COG3DBF			912h	CM5NSEL										
813h	COG3CON0			913h	CM5PSEL										
814h	COG3CON1			914h	CM6CON0										
815h	COG3RIS0		Unimplemented	915h	CM6CON1		Unimplemented		Unimplemented		Unimplemented		Unimplemented		Unimplemented
816h	COG3RIS1		Read as '0'	916h	CM6NSEL		Read as '0'		Read as '0'		Read as '0'		Read as '0'		Read as '0'
817h	COG3RSIM0			917h	CM6PSEL										
818h	COG3RSIM1			918h											
819h	COG3FIS0														
81Ah	COG3FIS1				Unimplemented										
81Bh	COG3FSIM0														
81Ch	COG3FSIM1 COG3ASD0				Read as '0'										
81Dh 81Eh	COG3ASD0 COG3ASD1														
81Fh	COG3STR	89Fh		91Fh		99Fh		A1Fh		A9Fh		B1Fh		B9Fh	
820h	0000011	8A0h		920h		9A0h		A20h		AA0h		B20h		BA0h	
020	General Purpose Register 48 Bytes		General Purpose Register 48 Bytes	02011	General Purpose Register 48 Bytes	0, 10,11	General Purpose Register 48 Bytes	,	General Purpose Register 48 Bytes		General Purpose Register 48 Bytes	52011	General Purpose Register 48 Bytes		General Purpose Register 48 Bytes
86Fh		8EFh		96Fh		9EFh		A6Fh		AEFh	, ,	B6Fh		BEFh	
870h		8F0h		90F11 970h		9F0h		A70h		AF0h		B70h		BF0h	
07011	Accesses 70h – 7Fh	51 011	Accesses 70h – 7Fh	07011	Accesses 70h – 7Fh	51 011	Accesses 70h – 7Fh	, (, 011	Accesses 70h – 7Fh	74 011	Accesses 70h – 7Fh	5,01	Accesses 70h – 7Fh	5, 01	Accesses 70h – 7Fh
87Fh		8FFh		97Fh		9FFh		A7Fh		AFFh		B7Fh		BFFh	

Legend: = Unimplemented data memory locations, read as '0'.

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TABLE 3-9: PIC16(L)F1777 MEMORY MAP, BANK 16-23

	BANK 16		BANK 17		BANK 18		BANK 19		BANK 20		BANK 21		BANK 22		BANK 23
800h		880h		900h		980h		A00h		A80h		B00h		B80h	
	Core Registers (Table 3-2)		Core Registers (Table 3-2)		Core Registers (Table 3-2)		Core Registers (Table 3-2)		Core Registers (Table 3-2)		Core Registers (Table 3-2)		Core Registers (Table 3-2)		Core Registers (Table 3-2)
80Bh	(12016-0-2)	88Bh	(10010-2)	90Bh	(10010-2)	98Bh	(10010-2)	A0Bh	(10010-0-2)	A8Bh	(Table 3-2)	B0Bh	(Table 3-2)	B8Bh	(14016-0-2)
80Ch	_	88Ch	_	90Ch	CM4CON0	98Ch		A0Ch		A8Ch		B0Ch		B8Ch	
80Dh	COG3PHR	00011	COG4PHR	90Dh	CM4CON1	00011		7.0011		/ 10 011		Doon		Boon	
80Eh	COG3PHF		COG4PHF	90Eh	CM4NSEL										
80Fh	COG3BLKR		COG4BLKR	90Fh	CM4PSEL										
810h	COG3BLKF		COG4BLKF	910h	CM5CON0										
811h	COG3DBR		COG4DBR	911h	CM5CON1										
812h	COG3DBF		COG4DBF	912h	CM5NSEL										
813h	COG3CON0		COG4CON0	913h	CM5PSEL										
814h	COG3CON1		COG4CON1	914h	CM6CON0										
815h	COG3RIS0		COG4RIS0	915h	CM6CON1										
816h	COG3RIS1		COG4RIS1	916h	CM6NSEL										
817h	COG3RSIM0		COG4RSIM0	917h	CM6PSEL		Unimplemented		Unimplemented		Unimplemented		Unimplemented		Unimplemented
818h	COG3RSIM1		COG4RSIM1	918h	CM7CON0		Read as '0'		Read as '0'		Read as '0'		Read as '0'		Read as '0'
819h	COG3FIS0		COG4FIS0		CM7CON1										
81Ah	COG3FIS1		COG4FIS1		CM7NSEL										
81Bh	COG3FSIM0		COG4FSIM0		CM7PSEL										
81Ch	COG3FSIM1		COG4FSIM1		CM8CON0										
81Dh	COG3ASD0		COG4ASD0		CM8CON1										
81Eh	COG3ASD1		COG4ASD1		CM8NSEL										
81Fh	COG3STR	89Fh	COG4STR	91Fh	CM8PSEL										
820h		8A0h		920h											
	Unimplemented		Unimplemented		Unimplemented										
	Read as '0'		Read as '0'		Read as '0'										
86Fh		8EFh		96Fh		9EFh		A6Fh		AEFh		B6Fh		BEFh	
870h		8F0h		970h		9F0h		A70h		AF0h		B70h		BF0h	
	Accesses		Accesses 70h – 7Fh		Accesses		Accesses		Accesses		Accesses		Accesses		Accesses
	70h – 7Fh		/UN – /FN		70h – 7Fh		70h – 7Fh		70h – 7Fh		70h – 7Fh		70h – 7Fh		70h – 7Fh
87Fh		8FFh		97Fh		9FFh		A7Fh		AFFh		B7Fh		BFFh	

Legend: = Unimplemented data memory locations, read as '0'.

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TABLE 3-10: PIC16(L)F1779 MEMORY MAP, BANK 16-23

		•••(I
	BANK 16		BANK 17		BANK 18		BANK 19		BANK 20		BANK 21		BANK 22		BANK 23
800h		880h		900h		980h		A00h		A80h		B00h		B80h	
	Core Registers (Table 3-2)		Core Registers (Table 3-2)		Core Registers (Table 3-2)		Core Registers (Table 3-2)		Core Registers (Table 3-2)		Core Registers (Table 3-2)		Core Registers (Table 3-2)		Core Registers (Table 3-2)
80Bh	(10010-0-2)	88Bh	(10010-0-2)	90Bh	(10010 0 2)	98Bh	(10010 0 2)	A0Bh	· · · ·	A8Bh	(10010 0 2)	B0Bh	(10010-0-2)	B8Bh	(10010 0 2)
80Ch	_	88Ch	_	90Ch	CM4CON0	98Ch		A0Ch		A8Ch		B0Ch		B8Ch	
80Dh	COG3PHR		COG4PHR	90Dh	CM4CON1			/ 10 011				200		200	
80Eh	COG3PHF		COG4PHF	90Eh	CM4NSEL										
80Fh	COG3BLKR		COG4BLKR	90Fh	CM4PSEL										
810h	COG3BLKF		COG4BLKF	910h	CM5CON0										
811h	COG3DBR		COG4DBR	911h	CM5CON1										
812h	COG3DBF		COG4DBF	912h	CM5NSEL										
813h	COG3CON0		COG4CON0	913h	CM5PSEL										
814h	COG3CON1		COG4CON1	914h	CM6CON0										
815h	COG3RIS0		COG4RIS0	915h	CM6CON1		Unimplemented		Unimplemented		Unimplemented		Unimplemented		Unimplemented
816h	COG3RIS1		COG4RIS1	916h	CM6NSEL		Read as '0'		Read as '0'		Read as '0'		Read as '0'		Read as '0'
817h	COG3RSIM0		COG4RSIM0	917h	CM6PSEL										
818h	COG3RSIM1		COG4RSIM1	918h	CM7CON0										
819h	COG3FIS0		COG4FIS0		CM7CON1										
81Ah	COG3FIS1		COG4FIS1		CM7NSEL										
81Bh	COG3FSIM0		COG4FSIM0		CM7PSEL										
81Ch	COG3FSIM1		COG4FSIM1		CM8CON0										
81Dh	COG3ASD0		COG4ASD0		CM8CON1										
81Eh	COG3ASD1		COG4ASD1		CM8NSEL										
81Fh	COG3STR	89Fh	COG4STR	91Fh	CM8PSEL	99Fh		A1Fh		A9Fh		B1Fh		B9Fh	
820h		8A0h		920h		9A0h		A20H		AA0h		B20h		BA0h	
	General Purpose		General Purpose		General Purpose		General Purpose		General Purpose		General Purpose		General Purpose		General Purpose
	Register 48 Bytes		Register 48 Bytes		Register 48 Bytes		Register 48 Bytes		Register 48 Bytes		Register 48 Bytes		Register 48 Bytes		Register 48 Bytes
86Fh		8EFh		96Fh		9EFh		A6Fh		AEFh		B6Fh		BEFh	
870h	A	8F0h	A	970h	A	9F0h	A	A70h	A	AF0h	A	B70h	A	BF0h	A
	Accesses 70h – 7Fh		Accesses 70h – 7Fh		Accesses 70h – 7Fh		Accesses 70h – 7Fh		Accesses 70h – 7Fh		Accesses 70h – 7Fh		Accesses 70h – 7Fh		Accesses 70h – 7Fh
87Fh	/	8FFh		97Fh		9FFh	-	A7Fh		AFFh	-	B7Fh		BFFh	/ 0.1. / 1.1.
07 FI		OFFI		SILLI		JITII		ATEI		AUTI		וודזע		DI'EII	

TABLE 3-11: PIC16(L)F1778 MEMORY MAP, BANK 24-31

	BANK 24		BANK 25		BANK 26		BANK 27		BANK 28		BANK 29		BANK 30		BANK 31
C00h	Core Registers (Table 3-2)	C80h	Core Registers (Table 3-2)	D00h	Core Registers (Table 3-2)	D80h	Core Registers (Table 3-2)	E00h	Core Registers (Table 3-2)	E80h	Core Registers (Table 3-2)	F00h	Core Registers (Table 3-2)	F80h	Core Registers (Table 3-2)
C0Bh		C8Bh		D0Bh		D8Bh		E0Bh		E8Bh		F0Bh		F8Bh	
C0Ch	_	C8Ch	_	D0Ch	_										
C0Dh	_	C8Dh	_	D0Dh											
C0Eh	_	C8Eh	_	D0Eh	_										
C0Fh	_	C8Fh	—	D0Fh	—										
C10h	_	C90h	—	D10h	—										
C11h	—	C91h	—	D11h	—										
C12h	_	C92h	—	D12h	—										
C13h	—	C93h	—	D13h											
C14h	—	C94h	—	D14h	—										
C15h	_	C95h	_	D15h	_										
C16h	_	C96h	_	D16h	—										
C17h	_	C97h	_	D17h	_		See Table 3-14		See Table 3-16						
C18h	—	C98h	_	D18h			for register map-								
C19h	_	C99h	_	D19h	—		ping details								
C1Ah	_	C9Ah	—	D1Ah	—										
C1Bh	_	C9Bh C9Ch	_	D1Bh	—										
C1Ch C1Dh		C9Ch C9Dh	—	D1Ch D1Dh											
C1Dh C1Eh		C9Dh C9Eh		D1Dh D1Eh											
C1Fh		C9Eh		D1Fh											
C1FII C20h		CA0h		D1FII D20h											
	General Purpose	CBFh	General Purpose Register 32 Bytes		Unimplemented										
	Register 80 Bytes	CC0h	Unimplemented Read as '0'		Read as '0'										
C6Fh		CEFh		D6Fh		DEFh		E6Fh		EEFh		F6Fh		FEFh	
C70h		CF0h		D70h		DF0h		E70h		EF0h		F70h		FF0h	
	Accesses		Accesses		Accesses		Accesses		Accesses		Accesses		Accesses		Accesses
	70h – 7Fh	0.55	70h – 7Fh		70h – 7Fh		70h – 7Fh		70h – 7Fh		70h – 7Fh		70h – 7Fh		70h – 7Fh
CFFh		CFFh		D7Fh		DFFh		E7Fh		EFFh		F7Fh		FFFh	

TABLE 3-12: PIC16(L)F1777 MEMORY MAP, BANK 24-31

	BANK 24		BANK 25		BANK 26		BANK 27		BANK 28		BANK 29		BANK 30		BANK 31
C00h	Core Registers (Table 3-2)	C80h	Core Registers (Table 3-2)	D00h	Core Registers (Table 3-2)	D80h	Core Registers (Table 3-2)	E00h	Core Registers (Table 3-2)	E80h	Core Registers (Table 3-2)	F00h	Core Registers (Table 3-2)	F80h	Core Registers (Table 3-2)
C0Bh		C8Bh		D0Bh		D8Bh		E0Bh		E8Bh		F0Bh		F8Bh	
C0Ch	_	C8Ch	_	D0Ch	_										
C0Dh	—	C8Dh	_	D0Dh	—										
C0Eh	—	C8Eh	—	D0Eh	_										
C0Fh	_	C8Fh	—	D0Fh	_										
C10h	_	C90h	_	D10h	—										
C11h	—	C91h	—	D11h	—										
C12h	—	C92h	—	D12h	_										
C13h	—	C93h	—	D13h	—										
C14h	—	C94h	—	D14h	—										
C15h		C95h	_	D15h	—										
C16h		C96h	—	D16h	—										
C17h	—	C97h	_	D17h	—		See Table 3-15		See Table 3-16						
C18h	_	C98h	_	D18h	_		for register map-								
C19h	_	C99h	—	D19h	_		ping details								
C1Ah		C9Ah C9Bh		D1Ah	 MD4CON0										
C1Bh C1Ch		C9Bn C9Ch		D1Bh D1Ch	MD4CON0 MD4CON1										
C1Dh		C9Dh	_	D1Dh	MD4CONT MD4SRC										
C1Eh		C9Eh		D1Eh	MD43RC MD4CARL										
C1Fh		C9Fh		D1Fh	MD4CARH										
C20h		CA0h		D20h	MD+CART										
				-											
	Unimplemented	CBFh	Unimplemented		Unimplemented										
	Read as '0'	CC0h	Read as '0'		Read as '0'										
C6Fh		CEFh		D6Fh		DEFh		E6Fh		EEFh		F6Fh		FEFh	
C70h		CF0h		D70h		DF0h		E70h		EF0h		F70h		FF0h	
	Accesses		Accesses		Accesses		Accesses		Accesses		Accesses		Accesses		Accesses
	70h – 7Fh		70h – 7Fh		70h – 7Fh		70h – 7Fh		70h – 7Fh		70h – 7Fh		70h – 7Fh		70h – 7Fh
CFFh		CFFh		D7Fh		DFFh		E7Fh		EFFh		F7Fh		FFFh	

TABLE 3-13: PIC16(L)F1779 MEMORY MAP, BANK 24-31

	BANK 24		BANK 25		BANK 26		BANK 27		BANK 28		BANK 29		BANK 30		BANK 31
C00h	Core Registers (Table 3-2)	C80h	Core Registers (Table 3-2)	D00h	Core Registers (Table 3-2)	D80h	Core Registers (Table 3-2)	E00h	Core Registers (Table 3-2)	E80h	Core Registers (Table 3-2)	F00h	Core Registers (Table 3-2)	F80h	Core Registers (Table 3-2)
C0Bh		C8Bh		D0Bh		D8Bh		E0Bh		E8Bh		F0Bh		F8Bh	
C0Ch	—	C8Ch	—	D0Ch	—										
C0Dh		C8Dh	—	D0Dh	—										
C0Eh		C8Eh	_	D0Eh											
C0Fh		C8Fh	_	D0Fh	—										
C10h		C90h	—	D10h	—										
C11h	_	C91h	_	D11h	—										
C12h		C92h	—	D12h											
C13h		C93h	—	D13h											
C14h C15h		C94h C95h		D14h D15h											
C150 C16h		C95h		D15h											
C100 C17h		C901		D17h											
C17h C18h		C9711		D1711 D18h			See Table 3-15		See Table 3-16						
C18h		C99h		D19h			for register map-								
C19h		C99fi C9Ah		D19h			ping details								
C1An C1Bh		C9An C9Bh		D1An D1Bh	 MD4CON0										
C1Bh C1Ch		C9Dh		D1Ch	MD4CON0										
C1Dh		C9Dh		D1Dh	MD4CON1 MD4SRC										
C1Eh		C9Eh		D1Eh	MD4GARL										
C1Fh	_	C9Fh	_	D1Fh	MD4CARH										
C20h		CA0h		D20h	MD+CART										
	General Purpose	CBFh	General Purpose Register 32 Bytes		Unimplemented										
	Register 80 Bytes	CC0h	Unimplemented Read as '0'		Read as '0'										
C6Fh		CEFh		D6Fh		DEFh		E6Fh		EEFh		F6Fh		FEFh	
C70h	Accesses 70h – 7Fh	CF0h	Accesses 70h – 7Fh	D70h	Accesses 70h – 7Fh	DF0h	Accesses 70h – 7Fh	E70h	Accesses 70h – 7Fh	EF0h	Accesses 70h – 7Fh	F70h	Accesses 70h – 7Fh	FF0h	Accesses 70h – 7Fh
CFFh		CFFh		D7Fh		DFFh		E7Fh		EFFh		F7Fh		FFFh	

TABLE 3-14: PIC16(L)F1778 MEMORY MAP, BANK 27-30

090h 091h 092h	PWMEN	E0Ch E0Dh	PPSLOCK INTPPS	E8Ch	_	F0Ch	_
8Eh 98Fh 90h 91h 92h		E0Dh	INTPPS				
08Fh 090h 091h 092h				E8Dh	—	F0Dh	_
D91h D92h		E0Eh	TOCKIPPS	E8Eh	_	F0Eh	_
D90h D91h D92h D93h	PWMLD	E0Fh	T1CKIPPS	E8Fh	—	F0Fh	CLCDATA
D92h	PWMOUT	E10h	T1GPPS	E90h	RA0PPS	F10h	CLC1CON
	PWM5PHL	E11h	T3CKIPPS	E91h	RA1PPS	F11h	CLC1POL
D93h	PWM5PHH	E12h	T3GPPS	E92h	RA2PPS	F12h	CLC1SEL0
	PWM5DCL	E13h	T5CKIPPS	E93h	RA3PPS	F13h	CLC1SEL1
D94h	PWM5DCH	E14h	T5GPPS	E94h	RA4PPS	F14h	CLC1SEL2
D95h	PWM5PRL	E15h	T2INPPS	E95h	RA5PPS	F15h	CLC1SEL3
D96h	PWM5PRH	E16h	T4INPPS	E96h	RA6PPS	F16h	CLC1GLS0
D97h	PWM50FL	E17h	T6INPPS	E97h	RA7PPS	F17h	CLC1GLS1
D98h	PWM50FH	E18h	T8INPPS	E98h	RB0PPS	F18h	CLC1GLS2
D99h	PWM5TMRL	E19h	CCP1PPS	E99h	RB1PPS	F19h	CLC1GLS3
D9Ah	PWM5TMRH	E1Ah	CCP2PPS	E9Ah	RB2PPS	F1Ah	CLC2CON
D9Bh	PWM5CON	E1Bh	CCP7PPS	E9Bh	RB3PPS	F1Bh	CLC2POL
D9Ch	PWM5INTE	E1Ch	_	E9Ch	RB4PPS	F1Ch	CLC2SEL0
D9Dh	PWM5INTF	E1Dh	COG1INPPS	E9Dh	RB5PPS	F1Dh	CLC2SEL1
D9Eh	PWM5CLKCON	E1Eh	COG2INPPS	E9Eh	RB6PPS	F1Eh	CLC2SEL2
D9Fh	PWM5LDCON	E1Fh	COG3INPPS	E9Fh	RB7PPS	F1Fh	CLC2SEL3
DA0h	PWM50FC0N	E20h	_	EA0h	RC0PPS	F20h	CLC2GLS0
DA1h	PWM6PHL	E21h	MD1CLPPS	EA1h	RC1PPS	F21h	CLC2GLS1
DA2h	PWM6PHH	E22h	MD1CHPPS	EA2h	RC2PPS	F22h	CLC2GLS2
DA3h	PWM6DCL	E23h	MD1MODPPS	EA3h	RC3PPS	F23h	CLC2GLS3
DA4h	PWM6DCH	E24h	MD2CLPPS	EA4h	RC4PPS	F24h	CLC3CON
DA5h	PWM6PRL	E25h	MD2CHPPS	EA5h	RC5PPS	F25h	CLC3POL
DA6h	PWM6PRH	E26h	MD2MODPPS	EA6h	RC6PPS	F26h	CLC3SEL0
DA7h	PWM60FL	E27h	MD3CLPPS	EA7h	RC7PPS	F27h	CLC3SEL1
DA8h	PWM60FH	E28h	MD3CHPPS	EA8h	_	F28h	CLC3SEL2
DA9h	PWM6TMRL	E29h	MD3MODPPS	EA9h	_	F29h	CLC3SEL3
DAAh	PWM6TMRH	E2Ah	_	EAAh	_	F2Ah	CLC3GLS0
DABh	PWM6CON	E2Bh	_	EABh	_	F2Bh	CLC3GLS1
DACh	PWM6INTE	E2Ch	_	EACh	_	F2Ch	CLC3GLS2
DADh	PWM6INTF	E2Dh	PRG1RPPS	EADh	_	F2Dh	CLC3GLS3
DAEh	PWM6CLKCON	E2Eh	PRG1FPPS	EAEh	_	F2Eh	CLC4CON
DAFh	PWM6LDCON	E2Fh	PRG2RPPS	EAFh		F2Fh	CLC4POL
DB0h	PWM60FC0N	E30h	PRG2FPPS	EB0h		F30h	CLC4SEL0
DB1h	PWM11PHL	E31h	PRG3FPPS	EB1h	_	F31h	CLC4SEL1
DB2h	PWM11PHH	E32h	PRG3RPPS	EB2h	_	F32h	CLC4SEL2
DB3h	PWM11DCL	E33h		EB3h	_	F33h	CLC4SEL3
DB4h	PWM11DCH	E34h		EB4h	_	F34h	CLC4GLS0
DB5h	PWM11PRL	E35h	CLCIN0PPS	EB5h		F35h	CLC4GLS1
DB6h	PWM11PRH	E36h	CLCIN1PPS	EB6h	_	F36h	CLC4GLS2
DB7h	PWM110FL	E37h	CLCIN2PPS	EB7h	_	F37h	CLC4GLS3
DB8h	PWM110FH	E38h	CLCIN3PPS	EB8h	_	F38h	_
DB9h	PWM11TMRL	E39h	ADCACTPPS	EB9h	_	F39h	_
DBAh	PWM11TMRH	E3Ah	SSPCLKPPS	EBAh	_	F3Ah	_
DBBh	PWM11CON	E3Bh	SSPDATPPS	EBBh	_	F3Bh	_
DBCh	PWM11INTE	E3Ch	SSPSSPPS	EBCh		F3Ch	_
DBDh	PWM11INTF	E3Dh	RXPPS	EBDh	_	F3Dh	_
DBEh	PWM11CLKCON	E3Eh	CKPPS	EBEh	_	F3Eh	_
DBFh	PWM11LDCON	E3Fh	_	EBFh	_	F3Fh	_
DC0h	PWM110FCON	E40h	_	EC0h		F40h	
DC1h							_
			—		—		
DEFh		E6Fh		EEFh		F6Fh	
egend:			memory locations,				

TABLE 3-15: PIC16(L)F1777/9 MEMORY MAP, BANK 27-30

LE 3-15:		1/////	MEMORY M	Аг, D	ANK 27-30		
	Bank 27		Bank 28		Bank 29		Bank 30
D8Ch	_	E0Ch	PPSLOCK	E8Ch		F0Ch	—
D8Dh	_	E0Dh	INTPPS	E8Dh	_	F0Dh	_
D8Eh	PWMEN	E0Eh	TOCKIPPS	E8Eh	_	F0Eh	—
D8Fh	PWMLD	E0Fh	T1CKIPPS	E8Fh	_	F0Fh	CLCDATA
D90h	PWMOUT	E10h	T1GPPS	E90h	RA0PPS	F10h	CLC1CON
D91h	PWM5PHL	E11h	T3CKIPPS	E91h	RA1PPS	F11h	CLC1POL
D92h	PWM5PHH	E12h	T3GPPS	E92h	RA2PPS	F12h	CLC1SEL0
D93h	PWM5DCL	E13h	T5CKIPPS	E93h	RA3PPS	F13h	CLC1SEL1
D94h	PWM5DCH	E14h	T5GPPS	E94h	RA4PPS	F14h	CLC1SEL2
D95h	PWM5PRL	E15h	T2INPPS	E95h	RA5PPS	F15h	CLC1SEL3
D96h	PWM5PRH	E16h	T4INPPS	E96h	RA6PPS	F16h	CLC1GLS0
D97h	PWM50FL	E17h	T6INPPS	E97h	RA7PPS	F17h	CLC1GLS1
D98h	PWM50FH	E18h	T8INPPS	E98h	RB0PPS	F18h	CLC1GLS2
D99h	PWM5TMRL	E19h	CCP1PPS	E99h	RB1PPS	F19h	CLC1GLS3
D9Ah	PWM5TMRH	E1Ah	CCP2PPS	E9Ah	RB2PPS	F1Ah	CLC2CON
D9Bh	PWM5CON	E1Bh	CCP7PPS	E9Bh	RB3PPS	F1Bh	CLC2POL
D9Ch	PWM5INTE	E1Ch	CCP8PPS	E9Ch	RB4PPS	F1Ch	CLC2SEL0
D9Dh	PWM5INTF	E1Dh	COG1INPPS	E9Dh	RB5PPS	F1Dh	CLC2SEL1
D9Eh	PWM5CLKCON	E1Eh	COG2INPPS	E9Eh	RB6PPS	F1Eh	CLC2SEL2
D9En D9Fh	PWM5LDCON	E1Fh	COG3INPPS	E9Eh	RB7PPS	F1Fh	CLC2SEL3
D9Fh DA0h	PWM50FC0N	E20h	COG4INPPS		RCOPPS		CLC2GLS0
DA01 DA1h	PWM50FC0N PWM6PHL	E20n E21h	MD1CLPPS	EA0h EA1h	RC1PPS	F20h F21h	CLC2GLS0 CLC2GLS1
DA2h DA3h	PWM6PHH PWM6DCL	E22h E23h	MD1CHPPS MD1MODPPS	EA2h EA3h	RC2PPS RC3PPS	F22h F23h	CLC2GLS2 CLC2GLS3
				EA3h			
DA4h	PWM6DCH	E24h	MD2CLPPS	EA4h	RC4PPS	F24h	CLC3CON
DA5h	PWM6PRL	E25h	MD2CHPPS	EA5h	RC5PPS	F25h	CLC3POL
DA6h	PWM6PRH	E26h	MD2MODPPS	EA6h	RC6PPS	F26h	CLC3SEL0
DA7h	PWM60FL	E27h	MD3CLPPS	EA7h	RC7PPS	F27h	CLC3SEL1
DA8h	PWM60FH	E28h	MD3CHPPS	EA8h	RD0PPS	F28h	CLC3SEL2
DA9h	PWM6TMRL	E29h	MD3MODPPS	EA9h	RD1PPS	F29h	CLC3SEL3
DAAh	PWM6TMRH	E2Ah	MD4CLPPS	EAAh	RD2PPS	F2Ah	CLC3GLS0
DABh	PWM6CON	E2Bh	MD4CHPPS	EABh	RD3PPS	F2Bh	CLC3GLS1
DACh	PWM6INTE	E2Ch	MD4MODPPS	EACh	RD4PPS	F2Ch	CLC3GLS2
DADh	PWM6INTF	E2Dh	PRG1RPPS	EADh	RD5PPS	F2Dh	CLC3GLS3
DAEh	PWM6CLKCON	E2Eh	PRG1FPPS	EAEh	RD6PPS	F2Eh	CLC4CON
DAFh	PWM6LDCON	E2Fh	PRG2RPPS	EAFh	RD7PPS	F2Fh	CLC4POL
DB0h	PWM60FC0N	E30h	PRG2FPPS	EB0h	RE0PPS	F30h	CLC4SEL0
DB1h	PWM11PHL	E31h	PRG3FPPS	EB1h	RE1PPS	F31h	CLC4SEL1
DB2h	PWM11PHH	E32h	PRG3RPPS	EB2h	RE2PPS	F32h	CLC4SEL2
DB3h	PWM11DCL	E33h	PRG4FPPS	EB3h		F33h	CLC4SEL3
DB4h	PWM11DCH	E34h	PRG4RPPS	EB4h		F34h	CLC4GLS0
DB5h	PWM11PRL	E35h	CLCIN0PPS	EB5h		F35h	CLC4GLS1
DB6h	PWM11PRH	E36h	CLCIN1PPS	EB6h		F36h	CLC4GLS2
DB7h	PWM110FL	E37h	CLCIN2PPS	EB7h		F37h	CLC4GLS3
DB8h	PWM110FH	E38h	CLCIN3PPS	EB8h		F38h	
DB9h	PWM11TMRL	E39h	ADCACTPPS	EB9h		F39h	
DBAh	PWM11TMRH	E3Ah	SSPCLKPPS	EBAh		F3Ah	
DBBh	PWM11CON	E3Bh	SSPDATPPS	EBBh		F3Bh	
DBCh	PWM11INTE	E3Ch	SSPSSPPS	EBCh		F3Ch	
DBDh	PWM11INTF	E3Dh	RXPPS	EBDh		F3Dh	
DBEh	PWM11CLKCON	E3Eh	CKPPS	EBEh		F3Eh	
DBFh	PWM11LDCON	E3Fh		EBFh		F3Fh	
DC0h	PWM110FCON	E40h		EC0h		F40h	
DC1h	PWM12PHL	E31h		EB1h		F31h	
DC2h	PWM12PHH	E32h		EB2h		F32h	
DC3h	PWM12DCL	E33h		EB3h	_	F33h	
DC4h	PWM12DCH	E34h		EB4h		F34h	
DC5h	PWM12PRL	E35h		EB5h		F35h	
DC6h	PWM12PRH	E36h		EB6h		F36h	—
DC7h	PWM12OFL	E37h		EB7h		F37h	
DC8h	PWM120FH	E38h		EB8h		F38h	
DC9h	PWM12TMRL	E39h	_	EB9h		F39h	
DCAh	PWM12TMRH	E3Ah		EBAh		F3Ah	
DCBh	PWM12CON	E3Bh		EBBh		F3Bh	
DCCh	PWM12INTE	E3Ch		EBCh		F3Ch	
DCDh	PWM12INTF	E3Dh		EBDh		F3Dh	
DCEh	PWM12CLKCON	E3Eh		EBEh		F3Eh	
DCFh	PWM12LDCON	E3Fh		EBFh		F3Fh	
DCFh	PWM12LDCON	E3Fh		EBFh		F3Fh	
DD0h		E6Fh		EEFh		F6Fh	
	PWM12OFCON						
DD1h -	—	E6Fh		EEFh		F6Fh	
DEFh		E6Fh		EEFh		F6Fh	
Legend:	= Unimpleme	nted data	memory locations,	read as '	0',		

 TABLE 3-16:
 PIC16(L)F1777/8/9 MEMORY

 MAP, BANK 31

	Bank 31	
F8Ch	Unimplemented	
FE3h	Read as '0'	
FE4h	STATUS_SHAD	
FE5h	WREG_SHAD	
FE6h	BSR_SHAD	
FE7h	PCLATH_SHAD	
FE8h	FSR0L_SHAD	
FE9h	FSR0H_SHAD	
FEAh	FSR1L_SHAD	
FEBh	FSR1H_SHAD	
FECh	—	
FEDh	STKPTR	
FEEh	TOSL	
FEFh	TOSH	
Legend:	= Unimplemented da read as '0',	ata memory locations,

PIC16(L)F1777/8/9

3.4.5 CORE FUNCTION REGISTERS SUMMARY

The Core Function registers listed in Table 3-17 can be addressed from any Bank.

IADEE	ABLE 3-17. CORE FUNCTION REGISTERS SUMMARTS								<u>.</u>		
Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Bank (0-31										
x00h or x80h	INDF0		ng this locat /sical regist		ontents of F	SR0H/FSR	OL to addro	ess data n	nemory	XXXX XXXX	uuuu uuuu
x01h or x81h	INDF1		ng this locat /sical regist		ontents of F	SR1H/FSR	1L to addro	ess data n	nemory	XXXX XXXX	uuuu uuuu
x02h or x82h	PCL	Program	Program Counter (PC) Least Significant Byte							0000 0000	0000 0000
x03h or x83h	STATUS	_	_	_	TO	PD	Z	DC	С	1 1000	q quuu
x04h or x84h	FSR0L	Indirect D	ata Memor	y Address	0 Low Point	er				0000 0000	uuuu uuuu
x05h or x85h	FSR0H	Indirect D	ata Memor	y Address	0 High Poin	ter				0000 0000	0000 0000
x06h or x86h	FSR1L	Indirect D	ata Memor	y Address	1 Low Point	er				0000 0000	uuuu uuuu
x07h or x87h	FSR1H	Indirect D	ata Memor	y Address	1 High Poin	ter				0000 0000	0000 0000
x08h or x88h	BSR	_	_	_	BSR4	BSR3	BSR2	BSR1	BSR0	0 0000	0 0000
x09h or x89h	WREG	Working F	Working Register								uuuu uuuu
x0Ah or x8Ah	PCLATH	_	— Write Buffer for the upper 7 bits of the Program Counter							-000 0000	-000 0000
x0Bh or x8Bh	INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	0000 0000	0000 0000

TABLE 3-17: CORE FUNCTION REGISTERS SUMMARY⁽¹⁾

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: These registers can be addressed from any bank.

TABLE 3-18: S	SPECIAL FUNCTION REGISTER SUMMARY
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					Value on	Value on all					
Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	POR, BOR	other Resets
Bank	< 0										
00Ch	PORTA	RA7	RA6	RA5	RA4	RA3	RA2	RA1	RA0	XXXX XXXX	uuuu uuuu
00Dh	PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	XXXX XXXX	uuuu uuuu
00Eh	PORTC	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0	XXXX XXXX	uuuu uuuu
00Fh	PORTD ⁽³⁾	RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0	XXXX XXXX	uuuu uuuu
010h	PORTE	—			_	RE3	RE2 ⁽³⁾	RE1 ⁽³⁾	RE0 ⁽³⁾	xxxx	uuuu
011h	PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
012h	PIR2	OSFIF	C2IF	C1IF	COG1IF	BCL1IF	C4IF	C3IF	CCP2IF	0000 0000	0000 0000
013h	PIR3	—		COG2IF	ZCDIF	CLC4IF	CLC3IF	CLC2IF	CLC1IF	00 0000	00 0000
014h	PIR4	—	TMR8IF	TMR5GIF	TMR5IF	TMR3GIF	TMR3IF	TMR6IF	TRM4IF	-000 0000	-000 0000
015h	PIR5	CCP8IF ⁽³⁾	CCP7IF	COG4IF ⁽³⁾	COG3IF	C8IF ⁽³⁾	C7IF ⁽³⁾	C6IF	C5IF	000000	000000
016h	PIR6	—			_	PWM12IF ⁽³⁾	PWM11IF	PWM6IF	PWM5IF	0000	0000
017h	TMR0	Timer0 Module Re	egister							0000 0000	0000 0000
018h	TMR1L	Holding Register f	or the Least Signif	cant Byte of the 16	6-bit TMR1 Registe	er				XXXX XXXX	uuuu uuuu
019h	TMR1H	Holding Register f	or the Most Signifi	cant Byte of the 16	-bit TMR1 Registe	r				XXXX XXXX	uuuu uuuu
01Ah	T1CON	CS<	1:0>	CKPS	6<1:0>	OSCEN	SYNC	—	ON	0000 00-0	uuuu uu-u
01Bh	T1GCON	GE	GPOL	GTM	GSPM	GGO/DONE	GVAL	GSS	<1:0>	0000 0x00	uuuu uxuu
01Ch	TMR3L	Holding Register for the Least Significant Byte of the 16-bit TMR3 Register								XXXX XXXX	uuuu uuuu
01Dh	TMR3H	Holding Register f	or the Most Signific	cant Byte of the 16	-bit TMR3 Registe	r				XXXX XXXX	uuuu uuuu
01Eh	T3CON	CS<	1:0>	CKPS	6<1:0>	OSCEN	SYNC	_	ON	0000 00-0	uuuu uu-u
01Fh	T3GCON	GE GPOL GTM GSPM GGO/DONE GVAL GSS<1:0>							<1:0>	00x0 0x00	uuuu uxuu

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Bank	<u>د 1</u>										
08Ch	TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	1111 1111	1111 1111
08Dh	TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	1111 1111	1111 1111
08Eh	TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111
08Fh	TRISD ⁽³⁾	TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0	1111 1111	1111 1111
090h	TRISE	—	_	_	—	(1)	TRISE2 ⁽³⁾	TRISE1 ⁽³⁾	TRISE0 ⁽³⁾	1111	1111
091h	PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
092h	PIE2	OSFIE	C2IE	C1IE	COG1IE	BCL1IE	C4IE	C3IE	CCP2IE	0000 0000	0000 0000
093h	PIE3	—	—	COG2IE	ZCDIE	CLC4IE	CLC3IE	CLC2IE	CLC1IE	00 0000	00 0000
094h	PIE4	—	TMR8IE	TMR5GIE	TMR5IE	TMR3GIE	TMR3IE	TMR6IE	TRM4IE	-000 0000	-000 0000
095h	PIE5	CCP8IE ⁽³⁾	CCP7IE	COG4IE ⁽³⁾	COG3IE	C8IE ⁽³⁾	C7IE ⁽³⁾	C6IE	C5IE	0000 0000	0000 0000
096h	PIE6	—	_	_	—	PWM12IE ⁽³⁾	PWM11IE	PWM6IE	PWM5IE	0000	0000
097h	OPTION_REG	WPUEN	INTEDG	TMR0CS	TMR0SE	PSA		PS<2:0>		1111 1111	1111 1111
098h	PCON	STKOVF	STKUNF	_	RWDT	RMCLR	RI	POR	BOR	00-1 11qq	qq-q qquu
099h	WDTCON	—	—			WDTPS<4:0>			SWDTEN	01 0110	01 0110
09Ah	OSCTUNE	—	—			TUN	<5:0>			00 0000	00 0000
09Bh	OSCCON	SPLLEN		IRCF	IRCF<3:0>				<1:0>	0011 1-00	0011 1-00
09Ch	OSCSTAT	SOSCR	PLLR	OSTS	HFIOFR	HFIOFL	MFIOFR	LFIOFR	HFIOFS	0dd0 0d0d	dddd ddod
09Dh	BORCON	SBOREN	BORFS	_	_	_	_	_	BORRDY	10q	uuu
09Eh	FVRCON	FVREN	FVRRDY	TSEN	TSRNG	CDAFV	/R<1:0>	ADFVI	R<1:0>	0q00 0000	0q00 0000
09Fh	ZCD1CON	EN	_	OUT	POL	_	_	INTP	INTN	0-x000	0-x000

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Ban	k 2										
10Ch	LATA	LATA7	LATA6	LATA5	LATA4	LATA3	LATA2	LATA1	LATA0	xxxx xxxx	uuuu uuuu
10Dh	LATB	LATB7	LATB6	LATB5	LATB4	LATB3	LATB2	LATB1	LATB0	XXXX XXXX	uuuu uuuu
	LATC	LATC7	LATC6	LATC5	LATC4	LATC3	LATC2	LATC1	LATC0	XXXX XXXX	uuuu uuuu
	LATD ⁽³⁾	LATD7	LATD6	LATD5	LATD4	LATD3	LATD2	LATD1	LATD0	XXXX XXXX	uuuu uuuu
110h	LATE ⁽³⁾	—	_		—	—	LATE2	LATE1	LATE0	111	111
111h	CMOUT	MC8OUT ⁽³⁾	MC7OUT ⁽³⁾	MC6OUT	MC5OUT	MC4OUT	MC3OUT	MC2OUT	MC1OUT	0000 0000	0000 0000
112h	CM1CON0	ON	OUT	_	POL	ZLF	Reserved	HYS	SYNC	00-0 0100	00-0 0100
113h	CM1CON1	_	—	_	—	_	—	INTP	INTN	00	00
114h	CM1NSEL	_	—	_	—		NCH	<3:0>		0000	0000
115h	CM1PSEL	_	—	_	—		PCH	<3:0>		0000	0000
116h	CM2CON0	ON	OUT		POL	ZLF	Reserved	HYS	SYNC	00-0 0100	00-0 0100
117h	CM2CON1	—	—		—	—	—	INTP	INTN	00	00
118h	CM2NSEL	—	—		—		NCH	<3:0>		0000	0000
119h	CM2PSEL	_	—	_	—		PCH	<3:0>		0000	0000
11Ah	CM3CON0	ON	OUT	_	POL	ZLF	Reserved	HYS	SYNC	00-0 0100	00-0 0100
11Bh	CM3CON1	-	—	_	—	_	—	INTP	INTN	00	00
11Ch	CM3NSEL	_	—	-	—		NCH		0000	0000	
11Dh	CM3PSEL	_	—	-	—		PCH		0000	0000	
11Eh	—	Unimplemented								—	—
11Fh	—	Unimplemented								—	—

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets	
Ban	k 3											
18Ch	ANSELA	_	—	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0	11 1111	11 1111	
18Dh	ANSELB	_	—	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	11 1111	11 1111	
18Eh	ANSELC	ANSC7	ANSC6	ANSC5	ANSC4	ANSC3	ANSC2	_	—	1111 11	1111 11	
18Fh	ANSELD ⁽³⁾	ANSD7	ANSD6	ANSD5	ANSD4	ANSD3	ANSD2	ANSD1	ANSD0	1111 1111	1111 1111	
190h	ANSELE ⁽³⁾	—	—	—	—	_	ANSE2	ANSE1	ANSE0	111	111	
191h	PMADRL	Program Memory	rogram Memory Address Register Low Byte									
192h	PMADRH	_(1)	Program Memory	Address Register	High Byte					1000 0000	1000 0000	
193h	PMDATL	Program Memory	Read Data Regist	er Low Byte						XXXX XXXX	uuuu uuuu	
194h	PMDATH	_	—	Program Memory	Read Data Regist	er High Byte				xx xxxx	uu uuuu	
195h	PMCON1	(1)	CFGS	LWLO	FREE	WRERR	WREN	WR	RD	1000 x000	1000 q000	
196h	PMCON2	Program Memory	Control Register 2	2						0000 0000	0000 0000	
197h	VREGCON	_	—	—	—	—	—	VREGPM ⁽²⁾	Reserved	01	01	
198h	—	Unimplemented								_	—	
198h	—	Unimplemented								_	—	
199h	RC1REG	EUSART Receive	e Data Register							0000 0000	0000 0000	
19Ah	TX1REG	EUSART Transm	it Data Register							0000 0000	0000 0000	
19Bh	SP1BRGL				SP1BR	G<7:0>				0000 0000	0000 0000	
19Ch	SP1BRGH				SP1BR	G<15:8>				0000 0000	0000 0000	
19Dh	RC1STA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 0000	0000 0000	
19Eh	TX1STA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	0000 0010	0000 0010	
19Fh	BAUD1CON	ABDOVF	RCIDL	—	SCKP	BRG16	_	WUE	ABDEN	01-0 0-00	01-0 0-00	
199h 19Fh		Unimplemented								_	_	

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Ban	k 4										
20Ch	WPUA	WPUA7	WPUA6	WPUA5	WPUA4	WPUA3	WPUA2	WPUA1	WPUA0	1111 1111	1111 1111
20Dh	WPUB	WPUB7	WPUB6	WPUB5	WPUB4	WPUB3	WPUB2	WPUB1	WPUB0	1111 1111	1111 1111
20Eh	WPUC	WPUC7	WPUC6	WPUC5	WPUC4	WPUC3	WPUC2	WPUC1	WPUC0	1111 1111	1111 1111
20Fh	WPUD ⁽³⁾	WPUD7	WPUD6	WPUD5	WPUD4	WPUD3	WPUD2	WPUD1	WPUD0	1111 1111	1111 1111
210h	WPUE	—	_	_	_	WPUE3	WPUE2 ⁽³⁾	WPUE1 ⁽³⁾	WPUE0 ⁽³⁾	1111	1111
211h	SSP1BUF	Synchronous Seri	ial Port Receive Bu	Iffer/Transmit Regi	ster					XXXX XXXX	uuuu uuuu
212h	SSP1ADD	ADD<7:0>						0000 0000	0000 0000		
213h	SSP1MSK				MSK	<7:0>				1111 1111	1111 1111
214h	SSP1STAT	SMP	CKE	D/Ā	Р	S	R/W	UA	BF	0000 0000	0000 0000
215h	SSP1CON1	WCOL	SSPOV	SSPEN	СКР		SSPN	1<3:0>		0000 0000	0000 0000
216h	SSP1CON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	0000 0000	0000 0000
217h	SSP1CON3	ACKTIM	PCIE	SCIE	BOEN	SDAHT	SBCDE	AHEN	DHEN	0000 0000	0000 0000
218h		Unimplemented									
21Ah	_	Unimplemented								_	_
21Bh	MD3CON0	EN	_	OUT	OPOL	—	_	_	BIT	0-000	0-000
21Ch	MD3CON1	—	_	CHPOL	CHSYNC	—	_	CLPOL	CLSYNC	0000	0000
21Dh	MD3SRC	_	—	—				0 0000	0 0000		
21Eh	MD3CARL	—	_	_				0 0000	0 0000		
21Fh	MD3CARH	—	_	_			CH<4:0>			0 0000	0 0000

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Bank	(5										
28Ch	ODCONA	ODA7	ODA6	ODA5	ODA4	ODA3	ODA2	ODA1	ODA0	0000 0000	0000 0000
28Dh	ODCONB	ODB7	ODB6	ODB5	ODB4	ODB3	ODB2	ODB1	ODB0	0000 0000	0000 0000
28Eh	ODCONC	ODC7	ODC6	ODC5	ODC4	ODC3	ODC2	ODC1	ODC0	0000 0000	0000 0000
28Fh	ODCOND ⁽³⁾	ODD7	ODD6	ODD5	ODD4	ODD3	ODD2	ODD1	ODD0	0000 0000	0000 0000
290h	ODCONE ⁽³⁾	_	-	—	_	—	ODE2	ODE1	ODE0	000	000
291h	CCPR1L	Capture/Compare	apture/Compare/PWM Register 1 (LSB)							XXXX XXXX	uuuu uuuu
292h	CCPR1H	Capture/Compare	oture/Compare/PWM Register 1 (MSB)								uuuu uuuu
293h	CCP1CON	EN	EN — OUT FMT MODE<3:0>								0-00 0000
294h	CCP1CAP	_	_	_	_		CTS	<3:0>		0000	0000
295h	CCPR2L	Capture/Compare	/PWM Register 2 (LSB)						xxxx xxxx	uuuu uuuu
296h	CCPR2H	Capture/Compare	/PWM Register 2 (MSB)						xxxx xxxx	uuuu uuuu
297h	CCP2CON	EN	_	OUT	FMT		MODE	E<3:0>		0-00 0000	0-00 0000
298h	CCP2CAP	_	_	_	_		CTS	<3:0>		0000	0000
299h	CCPR7L	Capture/Compare	/PWM Register 7 (LSB)						xxxx xxxx	uuuu uuuu
29Ah	CCPR7H	Capture/Compare	/PWM Register 7 (MSB)						xxxx xxxx	uuuu uuuu
29Bh	CCP7CON	EN	EN – OUT FMT MODE<3:0>								
29Ch	CCP7CAP	_	CTS<3:0>								0000
29Dh	_	Unimplemented								_	
29Eh	CCPTMRS1	C8TSEL	.<1:0> ⁽³⁾	C7TSE	L<1:0>	C2TSE	L<1:0>	C1TSI	EL<1:0>	00 0000	00 0000
29Fh	CCPTMRS2	P10TSE	P10TSEL<1:0> ⁽³⁾ P9TSEL<1:0> P4TSEL<1:0> P3TSEL<1:0>								00 0000

x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'. Legend:

Unimplemented, read as '1'.

Note 1: 2:

Unimplemented on PIC16LF1777/8/9.

IAD	LE 3-10: 31	CIAL FUN		STER SUM		TINUED)					
Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Ban	k 6										
30Ch	SLRCONA	SLRA7	SLRA6	SLRA5	SLRA4	SLRA3	SLRA2	SLRA1	SLRA0	1111 1111	1111 1111
30Dh	SLRCONB	SLRB7	SLRB6	SLRB5	SLRB4	SLRB3	SLRB2	SLRB1	SLRB0	1111 1111	1111 1111
30Eh	SLRCONC	SLRC7	SLRC6	SLRC5	SLRC4	SLRC3	SLRC2	SLRC1	SLRC0	1111 1111	1111 1111
30Fh	SLRCOND ⁽³⁾	SLRD7	SLRD6	SLRD5	SLRD4	SLRD3	SLRD0	1111 1111	1111 1111		
310h	SLRCONE ⁽³⁾		—	—	—	—	111	111			
311h	CCPR8L	Capture/Compare	e/PWM Register 8	(LSB)						xxxx xxxx	uuuu uuuu
312h	CCPR8H	Capture/Compare	e/PWM Register 8	(MSB)			XXXX XXXX	uuuu uuuu			
313h	CCP8CON	EN	—	OUT	FMT			0-00 0000	0-00 0000		
314h	CCP8CAP		—	—	_		CTS	<3:0>		0000	0000
315h	MD1CON0	EN	—	OUT	OPOL	_	—	—	BIT	0-000	0-000
316h	MD1CON1		—	CHPOL	CHSYNC	_	—	CLPOL	CLSYNC	0000	0000
317h	MD1SRC		—	—			MS<4:0>			0 0000	0 0000
318h	MD1CARL		—	—			CL<4:0>			0 0000	0 0000
319h	MD1CARH		—	—			CH<4:0>			0 0000	0 0000
31Ah	_	Unimplemented								—	—
31Bh	MD2CON0	EN	—	OUT	OPOL	—	BIT	0-000	0-000		
31Ch	MD2CON1		—	CHPOL	CHSYNC — — CLPOL CLSYNC					0000	0000
31Dh	MD2SRC		—	—	MS<4:0>					0 0000	0 0000
31Eh	MD2CARL		—	—	CL<4:0>					0 0000	0 0000
31Fh	MD2CARH		—	—				0 0000	0 0000		

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Banl	k7										
38Ch	INLVLA	INLVLA7	INLVLA6	INLVLA5	INLVLA4	INLVLA3	INLVLA2	INLVLA1	INLVLA0	1111 1111	1111 1111
38Dh	INLVLB	INLVLB7	INLVLB6	INLVLB5	INLVLB4	INLVLB3	INLVLB2	INLVLB1	INLVLB0	1111 1111	1111 1111
38Eh	INLVLC	INLVLC7	INLVLC6	INLVLC5	INLVLC4	INLVLC3	INLVLC2	INLVLC1	INLVLC0	1111 1111	1111 1111
38Fh	INLVLD ⁽³⁾	INLVLD7	INLVLD6	INLVLD5	INLVLD4	INLVLD3	INLVLD2	INLVLD1	INLVLD0	1111 1111	1111 1111
390h	INLVLE ⁽³⁾	—	_	_		INLVLE3	INLVLE2	INLVLE1	INLVLE0	1111	1111
391h	IOCAP	IOCAP7	IOCAP6	IOCAP5	IOCAP4	IOCAP3	IOCAP2	IOCAP1	IOCAP0	0000 0000	0000 0000
392h	IOCAN	IOCAN7	IOCAN6	IOCAN5	IOCAN4	IOCAN3	IOCAN2	IOCAN1	IOCAN0	0000 0000	0000 0000
393h	IOCAF	IOCAF7	IOCAF6	IOCAF5	IOCAF4	IOCAF3	IOCAF2	IOCAF1	IOCAF0	0000 0000	0000 0000
394h	IOCBP	IOCBP7	IOCBP6	IOCBP5	IOCBP4	IOCBP3	IOCBP2	IOCBP1	IOCBP0	0000 0000	0000 0000
395h	IOCBN	IOCBN7	IOCBN6	IOCBN5	IOCBN4	IOCBN3	IOCBN2	IOCBN1	IOCBN0	0000 0000	0000 0000
396h	IOCBF	IOCBF7	IOCBF6	IOCBF5	IOCBF4	IOCBF3	IOCBF2	IOCBF1	IOCBF0	0000 0000	0000 0000
397h	IOCCP	IOCCP7	IOCCP6	IOCCP5	IOCCP4	IOCCP3	IOCCP2	IOCCP1	IOCCP0	0000 0000	0000 0000
398h	IOCCN	IOCCN7	IOCCN6	IOCCN5	IOCCN4	IOCCN3	IOCCN2	IOCCN1	IOCCN0	0000 0000	0000 0000
399h	IOCCF	IOCCF7	IOCCF6	IOCCF5	IOCCF4	IOCCF3	IOCCF2	IOCCF1	IOCCF0	0000 0000	0000 0000
39Ah	—	Unimplemented								—	_
39Bh	—	Unimplemented								—	_
39Ch	—	Unimplemented								—	_
39Dh	IOCEP	_	_	_	_	IOCEP3	_	_	_	0	0
39Eh	IOCEN	_	_	—	_	IOCEN3	_	_	—	0	0
39Fh	IOCEF	_	_	—	_	IOCEF3	_	_	—	0	0

x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'. Legend:

Unimplemented, read as '1'.

Note 1:

Unimplemented on PIC16LF1777/8/9. 2:

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets			
Banl	< 8													
40Ch 40Ch	_	Unimplemented								_	—			
40Dh	HIDRVB	—	—		—	_	—	HIDB1	HIDB0	00	00			
40Eh	_	Unimplemented								—	—			
40Fh	TMR5L	Holding Register	for the Least Signif	icant Byte of the 16	6-bit TMR5 Registe	r				xxxx xxxx	uuuu uuuu			
410h	TMR5H	Holding Register	for the Most Signifi		XXXX XXXX	uuuu uuuu								
411h	T5CON	CS<	CS<1:0> CKPS<1:0> OSCEN SYNC ON											
412h	T5GCON	GE	GE GPOL GTM GSPM <u>GGO/</u> GVAL GSS<1:0>											
413h	T4TMR	Holding Register	for the 8-bit TMR4	Register						0000 0000	0000 0000			
414h	T4PR	TMR4 Period Reg	gister							1111 1111	1111 1111			
415h	T4CON	ON		CKPS<2:0>			OUTP	S<3:0>		0000 0000	0000 0000			
416h	T4HLT	PSYNC	CKPOL	CKSYNC			MODE<4:0>			0 0000	0 0000			
417h	T4CLKCON	—	_		—		CS<	:3:0>		0000	0000			
418h	T4RST	—	_				RSEL<4:0>			0 0000	0 0000			
419h	—	Unimplemented								—	—			
41Ah	T6TMR	Holding Register	for the 8-bit TMR6	Register						0000 0000	0000 0000			
41Bh	T6PR	TMR6 Period Reg	R6 Period Register											
41Ch	T6CON	ON		CKPS<2:0>			OUTP	S<3:0>		0000 0000	0000 0000			
41Dh	T6HLT	PSYNC	CKPOL	CKSYNC			MODE<4:0>			0 0000	0 0000			
41Eh	T6CLKCON	—	—	_	—		CS<	:3:0>		0000	0000			
41Fh	T6RST	—	—	_			RSEL<4:0>			0 0000	0 0000			

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

						- /					
Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Bank	¢ 9	_								_	
48Ch to 48Dh	_	Unimplemented								-	—
48Eh	ADRESL	ADC Result Regis	ster Low							XXXX XXXX	uuuu uuuu
48Fh	ADRESH	ADC Result Regis	ster High							XXXX XXXX	uuuu uuuu
490h	ADCON0			ADON	0000 0000	0000 0000					
491h	ADCON1	ADFM		F<1:0>	0000 -000	0000 -000					
492h	ADCON2	—	_		00 0000	00 0000					
493h	T2TMR	Holding Register f	or the 8-bit TMR2		0000 0000	0000 0000					
494h	T2PR	TMR2 Period Reg	jister							1111 1111	1111 1111
495h	T2CON	ON		CKPS<2:0>			OUTP	S<3:0>		0000 0000	0000 0000
496h	T2HLT	PSYNC	CKPOL	CKSYNC			MODE<4:0>			0000 0000	0000 0000
497h	T2CLKCON	_			—		CS<	:3:0>		0000	0000
498h	T2RST	_					RSEL<4:0>			0 0000	0 0000
499h		Unimplemented								—	_
49Ah	T8TMR	Holding Register f	or the 8-bit TMR8	Register						0000 0000	0000 0000
49Bh	T8PR	TMR8 Period Reg	jister		1111 1111	1111 1111					
49Ch	T8CON	ON			0000 0000	0000 0000					
49Dh	T8HLT	PSYNC	CKPOL	CKSYNC			MODE<4:0>			0000 0000	0000 0000
49Eh	T8CLKCON	_		_	—		CS<	:3:0>		0000	0000
49Fh	T8RST	_	_		0 0000	0 0000					

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets	
Banl	< 10											
50Ch 50Eh	_	Unimplemented								-	-	
50Fh	OPA1NCHS	_	_	_	_		NCH	<3:0>		0000	0000	
510h	OPA1PCHS	—	_	_	—		PCH	<3:0>		0000	0000	
511h	OPA1CON	EN	_	—	UG	—	ORPOL	ORM	<1:0>	00 -000	00 -000	
512h	OPA1ORS	—	_	—			ORS<4:0>			0 0000	0 0000	
513h	OPA2NCHS	_										
514h	OPA2PCHS	_	PCH<3:0>									
515h	OPA2CON	EN	-	_	UG	—	ORPOL	ORM	<1:0>	00 -000	00 -000	
516h	OPA2ORS	_	-	_			ORS<4:0>			0 0000	0 0000	
517h	OPA3NCHS	—	_	—	—		NCH	<3:0>		0000	0000	
518h	OPA3PCHS	—	_	—	—		PCH	<3:0>		0000	0000	
519h	OPA3CON	EN	SP	—	UG	—	ORPOL	ORM	<1:0>	00-0 -000	00-0 -000	
51Ah	OPA3ORS	—	_	—			ORS<4:0>			0 0000	0 0000	
51Bh	OPA4NCHS ⁽³⁾	_		—	—		NCH	<3:0>		0000	0000	
51Ch	OPA4PCHS ⁽³⁾	PCH<3:0>									0000	
51Dh	OPA4CON ⁽³⁾	EN SP — UG — ORPOL ORM<1:0>									00-0 -000	
51Eh	OPA4ORS ⁽³⁾	—	_	_			ORS<4:0>			0 0000	0 0000	
51Fh	—	Unimplemented	Unimplemented									

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Bank	c 11										
58Ch	_	Unimplemented								—	—
58Dh	DACLD	—	—	DAC6LD ⁽³⁾	DAC5LD	—		DAC2LD	DAC1LD	0000	0000
58Eh	DAC1CON0	EN	FM	OE1	OE2	PSS	<1:0>	NSS	<1:0>	0000 0000	0000 0000
58Fh	DAC1REFL				REF	REF<7:0>					0000 0000
590h	DAC1REFH				REF<15:8> OE2 PSS<1:0> NSS<1:0>					00000 00000	0000 0000
591h	DAC2CON0	EN	FM	OE1	OE2	PSS	0000 0000	0000 0000			
592h	DAC2REFL				REF<7:0>					00000 00000	0000 0000
593h	DAC2REFH				REF<15:8>					00000 00000	0000 0000
594h	DAC3CON0	EN	—	OE1	OE2	PSS	<1:0>	NSS	<1:0>	0-00 0000	0-00 0000
595h	DAC3REF	—	—				REF<4:0>			0 0000	0 0000
596h	DAC4CON0	EN	—	OE1	OE2	PSS	<1:0>	NSS	<1:0>	0-00 0000	0-00 0000
597h	DAC4REF	—	—				REF<4:0>			0 0000	0000 0000
598h	DAC5CON0	EN	FM	OE1	OE2	PSS	<1:0>	NSS	<1:0>	0000 0000	0000 0000
599h	DAC5REFL				REF	<7:0>				00000 00000	0000 0000
59Ah	DAC5REFH				REF<	:15:8>				00000 00000	0000 0000
59Bh	DAC6CON0 ⁽³⁾	EN	FM	OE1	OE2 PSS<1:0> NSS<1:0>					0000 0000	0000 0000
59Ch	DAC6REFL ⁽³⁾				REF<7:0>				00000 00000	0000 0000	
59Dh	DAC6REFH ⁽³⁾				REF<15:8>					00000 00000	0000 0000
59Eh	DAC7CON0	EN	_	OE1	OE2 PSS<1:0> NSS<1:0>				<1:0>	0-00 0000	0-00 0000
59Fh	DAC7REF	—	—	-			REF<4:0>			0 0000	0000 0000

 Legend:
 x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

 Note
 1:
 Unimplemented, read as '1'.

 2:
 Unimplemented on PIC16LF1777/8/9.

Onimplemented on PIC16(L)F1778.
 Unimplemented on PIC16(L)F1778.

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Banl	k 12										
60Ch to 613h	_	Unimplemented								_	—
60Ch	DAC8CON0 ⁽³⁾	EN	_	OE1	OE2	PSS	<1:0>	NSS1	NSS0	0-00 0000	0-00 0000
60Dh	DAC8REF ⁽³⁾	—	_	-			REF<4:0>			0 0000	0000 0000
60Eh	PRG4RTSS ⁽³⁾	—	_	-			RTSS	6<3:0>		0000	0000
	PRG4FTSS ⁽³⁾	—					FTSS	<3:0>		0000	0000
610h	PRG4INS ⁽³⁾	—					INS<	<3:0>		0000	0000
611h	PRG4CON0 ⁽³⁾	EN	_	FEDG	REDG	MODE	E<1:0>	OS	GO	0-000 0000	0-00 0000
612h	PRG4CON1 ⁽³⁾	—				_	RDY	FPOL	RPOL	000	000
613h	PRG4CON2 ⁽³⁾	—					ISET<4:0>			0 0000	0 0000
614h	PWM3DCL	DC<	1:0>			_	_	—	—	xx	uu
615h	PWM3DCH				DC<	9:2>				XXXX XXXX	uuuu uuuu
616h	PWM3CON	EN		OUT	POL	—	_	_	—	0-00	0-00
617h	PWM4DCL	DC<	1:0>			_	_	—	—	xx	uu
618h	PWM4DCH				DC<	9:2>				XXXX XXXX	uuuu uuuu
619h	PWM4CON	EN		OUT	POL	_	_	—	—	0-00	0-00
61Ah	PWM9DCL	DC<	1:0>	-	_	—	—	—	—	xx	uu
61Bh	PWM9DCH				DC<	9:2>				XXXX XXXX	uuuu uuuu
61Ch	PWM9CON	EN								0-00	0-00
61Dh	PWM10DCL ⁽³⁾	DC<	1:0>	-		—	—	—	—	xx	uu
61Eh	PWM10DCH ⁽³⁾	DC<9:2>							xxxx xxxx	uuuu uuuu	
61Fh	PWM10CON ⁽³⁾	EN	EN – OUT POL – – – –							0-00	0-00

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Bank	< 13										
68Ch	—	Unimplemented								—	—
68Dh	COG1PHR	—	_	COG Rising Edge	Phase Delay Cou	nt Register				00 0000	00 0000
68Eh	COG1PHF	—		COG Falling Edge	Phase Delay Cou	unt Register				00 0000	00 0000
68Fh	COG1BLKR	—		COG Rising Edge	Blanking Count R		00 0000	00 0000			
690h	COG1BLKF	—		COG Falling Edge	Blanking Count F		00 0000	00 0000			
691h	COG1DBR	—	_	COG Rising Edge	Dead-band Coun	00 0000	00 0000				
692h	COG1DBF	—		COG Falling Edge	Dead-band Coun	00 0000	00 0000				
693h	COG1CON0	EN	LD	—							
694h	COG1CON1	RDBS	FDBS	POLD POLC POLB POLA						00 0000	00 0000
695h	COG1RIS0	RIS7	RIS6	RIS5	RIS4	RIS3	RIS2	RIS1	RIS0	0000 0000	0000 0000
696h	COG1RIS1	RIS15	RIS14	RIS13	RIS12	RIS11	RIS10	RIS9	RIS8	0000 0000	0000 0000
697h	COG1RSIM0	RSIM7	RSIM6	RSIM5	RSIM4	RSIM3	RSIM2	RSIM1	RSIM0	0000 0000	0000 0000
698h	COG1RSIM1	RSIM15	RSIM14	RSIM13	RSIM12	RSIM11	RSIM10	RSIM9	RSIM8	0000 0000	0000 0000
699h	COG1FIS0	FIS7	FIS6	FIS5	FIS4	FIS3	FIS2	FIS1	FIS0	0000 0000	0000 0000
69Ah	COG1FIS1	FIS15	FIS14	FIS13	FIS12	FIS11	FIS10	FIS9	FIS8	0000 0000	0000 0000
69Bh	COG1FSIM0	FSIM7	FSIM6	FSIM5FSIM4FSIM3FSIM2FSIM1FSIM0						0000 0000	0000 0000
69Ch	COG1FSIM1	FSIM15	FSIM14	FSIM13 FSIM12 FSIM11 FSIM10 FSIM9 FSIM8						0000 0000	0000 0000
69Dh	COG1ASD0	ASE	ARSEN	ASDBI	0<1:0>	ASDA	C<1:0>			0001 01	0001 01
69Eh	COG1ASD1	AS7E	AS6E	AS5E AS4E AS3E AS2E AS1E AS0E					0000 0000	0000 0000	
69Fh	COG1STR	SDATD	SDATC	SDATB SDATA STRD STRC STRB STRA						0000 0000	0000 0000

x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'. Legend:

Note 1: Unimplemented, read as '1'.

2:

Unimplemented on PIC16LF1777/8/9. 3: Unimplemented on PIC16(L)F1778.

					•	,				
Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR
Banl	k 14									
70Ch	—	Unimplemented								—
70Dh	COG2PHR	—		COG Rising Edge	Phase Delay Cou	int Register				00 0000
70Eh	COG2PHF	—		COG Falling Edge	e Phase Delay Cou	unt Register				00 0000
70Fh	COG2BLKR	—		COG Rising Edge	Blanking Count R	legister				00 0000
710h	COG2BLKF	—		COG Falling Edge	e Blanking Count F	Register				00 0000
711h	COG2DBR	—		COG Rising Edge	Dead-band Coun	t Register				00 0000
712h	COG2DBF	—		COG Falling Edge	e Dead-band Coun	it Register				00 0000
713h	COG2CON0	EN	LD	_	CS<	:1:0>		MD<2:0>		00-0 0000

_

RIS4

RIS12

RSIM4

RSIM12

FIS4

FIS12

FSIM4

FSIM12

AS4E

SDATA

POLD

RIS3

RIS11

RSIM3

RSIM11

FIS3

FIS11

FSIM3

FSIM11

AS3E

STRD

ASDAC<1:0>

POLC

RIS2

RIS10

RSIM2

RSIM10

FIS2

FIS10

FSIM2

FSIM10

AS2E

STRC

POLB

RIS1

RIS9

RSIM1

RSIM9

FIS1

FIS9

FSIM1

FSIM9

_

AS1E

STRB

POLA

RIS0

RIS8

RSIM0

RSIM8

FIS0

FIS8

FSIM0

FSIM8

_

AS0E

STRA

00-- 0000

0000 0000

0000 0000

0000 0000

0000 0000

0000 0000

0000 0000

0000 0000

0000 0000

0001 01--

0000 0000

0000 0000

TABLE 3-18: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

FDBS

RIS6

RIS14

RSIM6

RSIM14

FIS6

FIS14

FSIM6

FSIM14

ARSEN

AS6E

SDATC

_

RIS5

RIS13

RSIM5

RSIM13

FIS5

FIS13

FSIM5

FSIM13

AS5E

SDATB

ASDBD<1:0>

Legend: x = unknown, u = unchanged, q = value depends on condition, -= unimplemented, read as '0', r = reserved.Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

RDBS

RIS7

RIS15

RSIM7

RSIM15

FIS7

FIS15

FSIM7

FSIM15

ASE

AS7E

SDATD

3: Unimplemented on PIC16(L)F1778.

714h

715h

716h

717h

718h

719h

71Ah

71Bh

71Ch

71Dh

71Eh

COG2CON1

COG2RIS0

COG2RIS1

COG2RSIM0

COG2RSIM1

COG2FIS0

COG2FIS1

COG2FSIM0

COG2FSIM1

COG2ASD0

COG2ASD1

71Fh COG2STR

Value on all other Resets

--00 0000 --00 0000 --00 0000 --00 0000 --00 0000 00-0 0000

00-- 0000

0000 0000

0000 0000

0000 0000

0000 0000

0000 0000

0000 0000

0000 0000

0000 0000

0001 01--

0000 0000

0000 0000

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Bank	k 15										
78Ch											
 78Dh	—	Unimplemented								-	_
78Eh	PRG1RTSS	—	—	—	—		RTSS	6<3:0>		0000	0000
78Fh	PRG1FTSS	_	_	_	—		FTSS	<3:0>		0000	0000
790h	PRG1INS	—	_		—			0000	0000		
791h	PRG1CON0	EN	_	FEDG	REDG	MODE	GO	0-000 0000	0-00 0000		
792h	PRG1CON1	—	_		_	000	000				
793h	PRG1CON2	—	—	_			ISET<4:0>			0 0000	0 0000
794h	PRG2RTSS	—	—	_	—		RTSS	6<3:0>		0000	0000
795h	PRG2FTSS	—	—	_	—		FTSS	<3:0>		0000	0000
796h	PRG2INS	—	—	_	—		INS<	<3:0>		0000	0000
797h	PRG2CON0	EN	—	FEDG	REDG	MODE	<1:0>	OS	GO	0-000 0000	0-00 0000
798h	PRG2CON1	—	—	_	—	—	RDY	FPOL	RPOL	000	000
799h	PRG2CON2	—	—	_			ISET<4:0>			0 0000	0 0000
79Ah	PRG3RTSS	—	—	_	—		RTSS	6<3:0>		0000	0000
79Bh	PRG3FTSS	—	—	_	—		FTSS	<3:0>		0000	0000
79Ch	PRG3INS	—	_		_		INS	<3:0>		0000	0000
79Dh	PRG3CON0	EN	_	FEDG	REDG	MODE	GO	0-000 0000	0-00 0000		
79Eh	PRG3CON1	—	_		_	—	RPOL	000	000		
79Fh	PRG3CON2	_	-	_				0 0000	0 0000		

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Unimplemented, read as '1'. Note 1:

Unimplemented on PIC16LF1777/8/9. 2:

	LE 5-10. 51	LOIALION									
Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Bank	c 16										
80Ch	—	Unimplemented								—	—
80Dh	COG3PHR	—		COG Rising Edge	e Phase Delay Coι	int Register				00 0000	00 0000
80Eh	COG3PHF	_	_	COG Falling Edge	e Phase Delay Cou	unt Register				00 0000	00 0000
80Fh	COG3BLKR	_	_	COG Rising Edge	Blanking Count R	Register				00 0000	00 0000
810h	COG3BLKF	_		COG Falling Edge	e Blanking Count F	Register				00 0000	00 0000
811h	COG3DBR	_		COG Rising Edge	e Dead-band Coun	t Register				00 0000	00 0000
812h	COG3DBF	_		COG Falling Edge	e Dead-band Cour	nt Register	•			00 0000	00 0000
813h	COG3CON0	EN	LD	_	CS<	:1:0>		MD<2:0>		00-0 0000	00-0 0000
814h	COG3CON1	RDBS	FDBS	—		POLD	POLC	POLB	POLA	00 0000	00 0000
815h	COG3RIS0	RIS7	RIS6	RIS5	RIS4	RIS3	RIS2	RIS1	RIS0	0000 0000	0000 0000
816h	COG3RIS1	RIS15	RIS14	RIS13	RIS12	RIS11	RIS10	RIS9	RIS8	0000 0000	0000 0000
817h	COG3RSIM0	RSIM7	RSIM6	RSIM5	RSIM4	RSIM3	RSIM2	RSIM1	RSIM0	0000 0000	0000 0000
818h	COG3RSIM1	RSIM15	RSIM14	RSIM13	RSIM12	RSIM11	RSIM10	RSIM9	RSIM8	0000 0000	0000 0000
819h	COG3FIS0	FIS7	FIS6	FIS5	FIS4	FIS3	FIS2	FIS1	FIS0	0000 0000	0000 0000
81Ah	COG3FIS1	FIS15	FIS14	FIS13	FIS12	FIS11	FIS10	FIS9	FIS8	0000 0000	0000 0000
81Bh	COG3FSIM0	FSIM7	FSIM6	FSIM5	FSIM4	FSIM3	FSIM2	FSIM1	FSIM0	0000 0000	0000 0000
81Ch	COG3FSIM1	FSIM15	FSIM14	FSIM13	FSIM12	FSIM11	FSIM10	FSIM9	FSIM8	0000 0000	0000 0000
81Dh	COG3ASD0	ASE	ARSEN	ASDBI	D<1:0>	ASDA	C<1:0>	—	—	0001 01	0001 01
81Eh	COG3ASD1	AS7E	AS6E	AS5E	AS4E	AS3E	AS2E	AS1E	AS0E	0000 0000	0000 0000
81Fh	COG3STR	SDATD	SDATC	SDATB	SDATA	STRD	STRC	STRB	STRA	0000 0000	0000 0000

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Banl	< 17										
88Ch	_	Unimplemented								_	_
88Dh	COG4PHR ⁽³⁾	—	_	COG Rising Edge	Phase Delay Cou	nt Register				00 0000	00 0000
88Eh	COG4PHF ⁽³⁾	—	_	COG Falling Edge	e Phase Delay Cou	int Register				00 0000	00 0000
88Fh	COG4BLKR ⁽³⁾	—	_	COG Rising Edge	Blanking Count R	egister				00 0000	00 0000
890h	COG4BLKF ⁽³⁾	—	_	COG Falling Edge	e Blanking Count F	egister				00 0000	00 0000
891h	COG4DBR ⁽³⁾	—	_	COG Rising Edge	Dead-band Coun	Register				00 0000	00 0000
892h	COG4DBF ⁽³⁾	—	_	COG Falling Edge	e Dead-band Coun	t Register				00 0000	00 0000
893h	COG4CON0 ⁽³⁾	EN	LD	—	CS<	1:0>		MD<2:0>		00-0 0000	00-0 0000
894h	COG4CON1 ⁽³⁾	RDBS	FDBS	—	_	POLD	POLC	POLB	POLA	00 0000	00 0000
895h	COG4RIS0(3)	RIS7	RIS6	RIS5	RIS4	RIS3	RIS2	RIS1	RIS0	0000 0000	0000 0000
896h	COG4RIS1 ⁽³⁾	RIS15	RIS14	RIS13	RIS12	RIS11	RIS10	RIS9	RIS8	0000 0000	0000 0000
897h	COG4RSIM0(3)	RSIM7	RSIM6	RSIM5	RSIM4	RSIM3	RSIM2	RSIM1	RSIM0	0000 0000	0000 0000
898h	COG4RSIM1 ⁽³⁾	RSIM15	RSIM14	RSIM13	RSIM12	RSIM11	RSIM10	RSIM9	RSIM8	0000 0000	0000 0000
899h	COG4FIS0 ⁽³⁾	FIS7	FIS6	FIS5	FIS4	FIS3	FIS2	FIS1	FIS0	0000 0000	0000 0000
89Ah	COG4FIS1 ⁽³⁾	FIS15	FIS14	FIS13	FIS12	FIS11	FIS10	FIS9	FIS8	0000 0000	0000 0000
89Bh	COG4FSIM0(3)	FSIM7	FSIM6	FSIM5	FSIM4	FSIM3	FSIM2	FSIM1	FSIM0	0000 0000	0000 0000
89Ch	COG4FSIM1 ⁽³⁾	FSIM15	FSIM14	FSIM13	FSIM12	FSIM11	FSIM10	FSIM9	FSIM8	0000 0000	0000 0000
89Dh	COG4ASD0(3)	ASE	ARSEN	ASDBI	D<1:0>	ASDA	C<1:0>	_	—	0001 01	0001 01
89Eh	COG4ASD1 ⁽³⁾	AS7E	AS6E	AS5E	AS4E	AS3E	AS2E	AS1E	AS0E	0000 0000	0000 0000
89Fh	COG4STR ⁽³⁾	SDATD	SDATC	SDATB	SDATA	STRD	STRC	STRB	STRA	0000 0000	0000 0000

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

IAD	LE 3-10: 3P			STER SUM		TINUED					
Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Ban	k 18										
90Ch	CM4CON0	ON	OUT	_	POL	ZLF	Reserved	HYS	SYNC	00-0 0100	00-0 0100
90Dh	CM4CON1	_	_	_	—	_	—	INTP	INTN	00	00
90Eh	CM4NSEL	—	—	—	—		NCH	<3:0>		0000	0000
90Fh	CM4PSEL	—	—	—	_		PCH	<3:0>		0000	0000
910h	CM5CON0	ON	OUT	—	POL	ZLF	Reserved	HYS	SYNC	00-0 0100	00-0 0100
911h	CM5CON1	—	—	—	_	—	—	INTP	INTN	00	00
912h	CM5NSEL	—	—	—	_		NCH	<3:0>		0000	0000
913h	CM5PSEL	—	—	—	—		PCH	<3:0>		0000	0000
914h	CM6CON0	ON	OUT	—	POL	ZLF	Reserved	HYS	SYNC	00-0 0100	00-0 0100
915h	CM6CON1	—	—	—	_	—	—	INTP	INTN	00	00
916h	CM6NSEL	—	—	—	_		NCH	<3:0>		0000	0000
917h	CM6PSEL	—	—	—	—		PCH	<3:0>		0000	0000
918h	CM7CON0 ⁽³⁾	ON	OUT	—	POL	ZLF	Reserved	HYS	SYNC	00-0 0100	00-0 0100
919h	CM7CON1 ⁽³⁾	—	—	—	_	—	—	INTP	INTN	00	00
91Ah	CM7NSEL ⁽³⁾	—	—	—	_		NCH	<3:0>		0000	0000
91Bh	CM7PSEL ⁽³⁾	—	—	—	_		PCH	<3:0>		0000	0000
91Ch	CM8CON0 ⁽³⁾	ON	OUT	—	POL	ZLF	Reserved	HYS	SYNC	00-0 0100	00-0 0100
91Dh	CM8CON1 ⁽³⁾	—	—	—	—	—	—	INTP	INTN	00	00
91Eh	CM8NSEL ⁽³⁾	—	—	—	—		NCH	<3:0>		0000	0000
91Fh	CM8PSEL ⁽³⁾	—	—	—	—		PCH	<3:0>		0000	0000

Bank 19-25

x0Ch/ x8Ch	_	Unimplemented	_	_
x1Fh/ x9Fh				

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Bank	c 26										
D0Ch		Unimplemented									
 D1Ah		Onimplemented								_	_
D1Bh	MD4CON0 ⁽³⁾	EN	-	OUT	OPOL	_	_	_	BIT	0-000	0-000
D1Ch	MD4CON1 ⁽³⁾	—	-	CHPOL	CHSYNC	_	_	CLPOL	CLSYNC	0000	0000
D1Dh	MD4SRC ⁽³⁾	—	-	—			MS<4:0>			0 0000	0 0000
D1Eh	MD4CARL ⁽³⁾	—	_	_			CL<4:0>			0 0000	0 0000
D1Fh	MD4CARH ⁽³⁾	—	_	_			CH<4:0>			0 0000	0 0000

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Banl	< 27										
D8Ch	_	Unimplemented								_	_
D8Dh	_	Unimplemented								_	_
D8Eh	PWMEN	_	_	_	_	MPWM12EN ⁽³⁾	MPWM11EN	MPWM6EN	MPWM5EN	0000	0000
D8Fh	PWMLD	_	_	_	_	MPWM12LD ⁽³⁾	MPWM11LD	MPWM6LD	MPWM5LD	0000	0000
D90h	PWMOUT	_	_	_	—	MPWM12OUT ⁽³⁾	MPWM11OUT	MPWM6OUT	MPWM5OUT	0000	0000
D91h	PWM5PHL				PH<	:7:0>				XXXX XXXX	uuuu uuuu
D92h	PWM5PHH				PH<	15:8>				XXXX XXXX	uuuu uuuu
D93h	PWM5DCL				DC<	:7:0>				XXXX XXXX	uuuu uuuu
D94h	PWM5DCH				DC<	15:8>				XXXX XXXX	uuuu uuuu
D95h	PWM5PRL				PR<	:7:0>				xxxx xxxx	uuuu uuuu
D96h	PWM5PRH				PR<	15:8>				xxxx xxxx	uuuu uuuu
D97h	PWM5OFL		PR<15:8> OF<7:0> OF<15:8>							XXXX XXXX	uuuu uuuu
D98h	PWM50FH		OF<15:8>							XXXX XXXX	uuuu uuuu
D99h	PWM5TMRL		OF<15:8> TMR<7:0>							0000 0000	0000 0000
D9Ah	PWM5TMRH									0000 0000	0000 0000
D9Bh	PWM5CON	EN	—	OUT	POL	MODE	<1:0>		—	0-00 00	0-00 00
D9Ch	PWM5INTE	—	_	—	—	OFIE	PHIE	DCIE	PRIE	0000	0000
D9Dh	PWM5INTF	—	_	—	—	OFIF	PHIF	DCIF	PRIF	0000	0000
D9Eh	PWM5CLKCON	—		PS<2:0>		—	—	CS<	:1:0>	-00000	-00000
D9Fh	PWM5LDCON	LDA	LDT		—	—	—	LDS	<1:0>	0000	0000
DA0h	PWM50FC0N	—	OFM	<1:0>	OFO	—	—	OFS	<1:0>	-00000	-00000
DA1h	PWM6PHL				PH<	:7:0>				XXXX XXXX	uuuu uuuu
DA2h	PWM6PHH				PH<	15:8>				XXXX XXXX	uuuu uuuu
DA3h	PWM6DCL				DC<	:7:0>				XXXX XXXX	uuuu uuuu
DA4h	PWM6DCH				DC<	15:8>				XXXX XXXX	uuuu uuuu
DA5h	PWM6PRL				PR<	:7:0>				XXXX XXXX	uuuu uuuu
DA6h	PWM6PRH		PR<15:8>							XXXX XXXX	uuuu uuuu
DA7h	PWM6OFL		OF<7:0>							XXXX XXXX	uuuu uuuu
DA8h	PWM6OFH				OF<	15:8>				XXXX XXXX	uuuu uuuu
DA9h	PWM6TMRL				TMR	<7:0>				0000 0000	0000 0000
DAAh	PWM6TMRH				TMR	<15:8>				0000 0000	0000 0000

 ${\bf x}$ = unknown, ${\bf u}$ = unchanged, ${\bf q}$ = value depends on condition, - = unimplemented, read as '0', ${\bf r}$ = reserved. Shaded locations are unimplemented, read as '0'. Legend:

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

IAD	-E 3-10. 3F			SIER SUIVI					-		
Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Bank	27 (Continued)										
DABh	PWM6CON	EN	—	OUT	POL	MODE	E<1:0>	—	—	0-00 00	0-00 00
DACh	PWM6INTE	—	_	—	—	OFIE	PHIE	DCIE	PRIE	0000	0000
DADh	PWM6INTF	—	—	—	—	OFIF	PHIF	DCIF	PRIF	0000	0000
DAEh	PWM6CLKCON	—		PS<2:0>		—	—	CS<	:1:0>	-00000	-00000
DAFh	PWM6LDCON	LDA	LDT	—	—	—	—	LDS	<1:0>	0000	0000
DB0h	PWM6OFCON	_	OFM	<1:0>	OFO			OFS	<1:0>	-00000	-00000
DB1h	PWM11PHL				PH<	:7:0>				XXXX XXXX	uuuu uuuu
DB2h	PWM11PHH				PH<	15:8>				XXXX XXXX	uuuu uuuu
DB3h	PWM11DCL				DC<	:7:0>				XXXX XXXX	uuuu uuuu
DB4h	PWM11DCH				DC<	15:8>				XXXX XXXX	uuuu uuuu
DB5h	PWM11PRL				PR<	:7:0>				XXXX XXXX	uuuu uuuu
DB6h	PWM11PRH				PR<	15:8>				XXXX XXXX	uuuu uuuu
DB7h	PWM110FL				OF<	:7:0>				xxxx xxxx	uuuu uuuu
DB8h	PWM110FH				OF<	15:8>				xxxx xxxx	uuuu uuuu
DB9h	PWM11TMRL				TMR	<7:0>				0000 0000	0000 0000
DBAh	PWM11TMRH				TMR	<15:8>				0000 0000	0000 0000
DBBh	PWM11CON	EN	_	OUT	POL	MODE	=<1:0>	_	_	0-00 00	0-00 00
DBCh	PWM11INTE	_	_	_	_	OFIE	PHIE	DCIE	PRIE	0000	0000
DBDh	PWM11INTF	_	_	_		OFIF	PHIF	DCIF	PRIF	0000	0000
DBEh	PWM11CLKCON	_		PS<2:0>	•		_	CS<	:1:0>	-00000	-00000
DBFh	PWM11LDCON	LDA	LDT	_	_		_	LDS	<1:0>	0000	0000
DC0h	PWM110FCON	—	OFM	<1:0>	OFO	—	—	OFS	<1:0>	-00000	-00000

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Bank	c 27 (Continued)										
DC1h	PWM12PHL ⁽³⁾				PH<	7:0>				XXXX XXXX	uuuu uuuu
DC2h	PWM12PHH ⁽³⁾				PH<	15:8>				XXXX XXXX	uuuu uuuu
DC3h	PWM12DCL ⁽³⁾				DC<	7:0>				XXXX XXXX	uuuu uuuu
DC4h	PWM12DCH ⁽³⁾				DC<	15:8>				XXXX XXXX	uuuu uuuu
DC5h	PWM12PRL ⁽³⁾				PR<	7:0>				XXXX XXXX	uuuu uuuu
DC6h	PWM12PRH ⁽³⁾				PR<	15:8>				XXXX XXXX	uuuu uuuu
DC7h	PWM12OFL ⁽³⁾					XXXX XXXX	uuuu uuuu				
DC8h	PWM12OFH ⁽³⁾					XXXX XXXX	uuuu uuuu				
DC9h	PWM12TMRL ⁽³⁾				0000 0000	0000 0000					
DCAh	PWM12TMRH ⁽³⁾				TMR<	:15:8>				0000 0000	0000 0000
DCBh	PWM12CON ⁽³⁾	EN	_	OUT	POL	MODI	E<1:0>	_	_	0-00 00	0-00 00
DCCh	PWM12INTE ⁽³⁾	_	_	_	_	OFIE	PHIE	DCIE	PRIE	0000	0000
DCDh	PWM12INTF ⁽³⁾	_	_	_	_	OFIF	PHIF	DCIF	PRIF	0000	0000
DCEh	PWM12CLKCON ⁽³⁾	_		PS<2:0>	·	_	—	CS<	:1:0>	-00000	-00000
DCFh	PWM12LDCON ⁽³⁾	LDA	LDT	<1:0>	0000	0000					
DD0h	PWM12OFCON ⁽³⁾	_	OFM	<1:0>	-00000	-00000					
DD1h to DEFh	_	Unimplemented			_	_					

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Bank	< 28	•	•		-			•	•	•	
E0Ch	_				Unimple		—	—			
 E0Bh											
E0Ch	PPSLOCK	-	—	—	—	PPSLOCKED	0	0			
E0Dh	INTPPS	_	_			INTPF	PS<5:0>			00 1000	uu uuuu
E0Eh	TOCKIPPS	_	_			T0CKIP	PS<5:0>			00 0100	uu uuuu
E0Fh	T1CKIPPS	_	_			T1CKIP	PS<5:0>			01 0000	uu uuuu
E10h	T1GPPS	_	_			T1GPF	PS<5:0>			00 1101	uu uuuu
E11h	T3CKIPPS	_	_			T3CKIP	PS<5:0>			01 0000	uu uuuu
E12h	T3GPPS	_	_			T3GPF	PS<5:0>			01 0000	uu uuuu
E13h	T5CKIPPS	_	_				01 0000	uu uuuu			
E14h	T5GPPS	_	_				00 1100	uu uuuu			
E15h	T2INPPS	_	_				01 0011	uu uuu			
E16h	T4INPPS	_	_			T4INPI	PS<5:0>			01 0101	uu uuuu
E17h	T6INPPS	_	_			T6INPI	PS<5:0>			01 0100	uu uuuu
E18h	T8INPPS	_	_			T8INPI	PS<5:0>			01 0010	uu uuuu
E19h	CCP1PPS	_	_			CCP1P	PS<5:0>			01 0010	uu uuuu
E1Ah	CCP2PPS	_	_			CCP2P	PS<5:0>			01 0001	uu uuuu
E1Bh	CCP7PPS	_	_			CCP7P	PS<5:0>			00 1101	uu uuuu
E1Ch	CCP8PPS ⁽³⁾	_	_			CCP8P	PS<5:0>			00 1101	uu uuuu
E1Dh	COGIN1PPS	_	_			COG1P	PS<5:0>			00 1000	uu uuu
E1Eh	COG2INPPS	_	_			COG2P	PS<5:0>			00 1001	uu uuu
E1Fh	COG3INPPS	_	_			COG3P	PS<5:0>			00 1010	uu uuu
E20h	COG4INPPS ⁽³⁾	_				COG4P	PS<5:0>			00 1010	uu uuu
E21h	MD1CLPPS	_				MD1CL	PPS<5:0>			00 0100	uu uuu
E22h	MD1CHPPS	_					00 0011	uu uuu			
E23h	MD1MODPPS	_	_				00 0101	uu uuuu			
E24h	MD2CLPPS	_	_			MD2CLF	PPS<5:0>			00 0100	uu uuuu
E25h	MD2CHPPS	_	_			MD2CHI	PPS<5:0>			00 0011	uu uuu
E26h	MD2MODPPS	—	_				00 0101	uu uuuu			

x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Unimplemented, read as '1'. Note 1:

> 2: Unimplemented on PIC16LF1777/8/9.

3: Unimplemented on PIC16(L)F1778. PIC16(L)F1777/8/9

	LL 3-10. 3F			STER SOM				•	•		
Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Ban	k 28 (Continued)										
E27h	MD3CLPPS	_	_			MD3CLF	PPS<5:0>			00 1100	uu uuuu
E28h	MD3CHPPS	—	—			MD3CH	PPS<5:0>			00 1011	uu uuuu
E29h	MD3MODPPS	—	—			MD3MOE)PPS<5:0>			00 0101	uu uuuu
E2Ah	MD4CLPPS ⁽³⁾	—	—			MD4CLF	PPS<5:0>			00 1100	uu uuuu
E2Bh	MD4CHPPS ⁽³⁾	—	—			MD4CH	PPS<5:0>			00 1011	uu uuuu
E2Ch	MD4MODPPS ⁽³⁾	—	—			MD4MOE)PPS<5:0>			00 0101	uu uuuu
E2Dh	PRG1RPPS	—	—			PRG1RF	PPS<5:0>			00 0100	uu uuuu
E2Eh	PRG1FPPS	—	—			PRG1FF	PPS<5:0>			00 0101	uu uuuu
E2Fh	PRG2RPPS	—	—	PRG2RPPS<5:0>01 0001 PRG2FPPS<5:0>01 0010						uu uuuu	
E30h	PRG2FPPS	—	—			PRG2FF	PPS<5:0>			01 0010	uu uuuu
E31h	PRG3RPPS	—	—			PRG3RF	PPS<5:0>			01 0100	uu uuuu
E32h	PRG3FPPS	—	—			PRG3FF	PPS<5:0>			01 0101	uu uuuu
	PRG4RPPS ⁽³⁾	—	—			PRG4RF	PPS<5:0>			01 0100	uu uuuu
E34h	PRG4FPPS ⁽³⁾	—	—			PRG4FF	PPS<5:0>			01 0101	uu uuuu
E35h	CLC1IN0PPS	—	—			CLCINO	PPS<5:0>			00 0000	uu uuuu
E36h	CLC1IN1PPS	—	—			CLCIN1	PPS<5:0>			00 0001	uu uuuu
E37h	CLC1IN2PPS	—	—	MD3CLPPS<5:0> MD3MODPPS<5:0> MD4CLPPS<5:0> MD4CLPPS<5:0> MD4CLPPS<5:0> MD4CLPPS<5:0> MD4CLPPS<5:0> MD4MODPPS<5:0> PRG1RPPS<5:0> PRG1PPS<5:0> PRG1PPS<5:0> PRG2RPPS<5:0> PRG3RPPS<5:0> PRG3PPS<5:0> PRG3PPS<5:0> PRG3PPS<5:0> PRG3PPS<5:0> PRG4PPS<5:0> PRG4PPS<5:0> PRG4PPS<5:0> PRG4PPS<5:0> CLCIN0PPS<5:0> CLCIN1PPS<5:0> CLCIN3PPS<5:0> CLCIN3PPS<5:0> SSPCLKPPS<5:0> SSPSDPPS<5:0> SSPSDPPS<5:0> RXPPS<5:0> CKPPS<5:0> CKPPS<5:0>			00 1110	uu uuuu			
E38h	CLC1IN3PPS	—	—			CLCIN3	PPS<5:0>			00 1111	uu uuuu
E39h	ADCACTPPS	—	—			ADCACT	PPS<5:0>			00 1100	uu uuuu
E3Ah	SSPCLKPPS	—	—			SSPCLK	PPS<5:0>			01 0011	uu uuuu
E3Bh	SSPDATPPS	—	—			SSPDAT	PPS<5:0>			01 0100	uu uuuu
E3Ch	SSPSSPPS	—	—			SSPSSF	PPS<5:0>			00 0101	uu uuuu
E3Dh	RXPPS		—				01 0111	uu uuuu			
E3Eh	CKPPS			CKPPS<5:0>						01 0110	uu uuuu
E3Fh	—				Unimple		—	—			
E6Fh				MD4MODPPS<5:0> PRG1RPPS<5:0> PRG1FPPS<5:0> PRG2FPPS<5:0> PRG2FPPS<5:0> PRG3RPPS<5:0> PRG3FPPS<5:0> PRG3FPPS<5:0> PRG3FPPS<5:0> PRG4FPPS<5:0> PRG4FPPS<5:0> CLCIN0PPS<5:0> CLCIN1PPS<5:0> CLCIN2PPS<5:0> CLCIN3PPS<5:0> ADCACTPPS<5:0> SSPDATPPS<5:0> SSPSAPPS<5:0> RXPPS<5:0> CKPPS<5:0>							

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

ТАВ		LOIAL I UNC									
Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Bank	29										
E8Ch E8Fh	_	Unimplemented								_	_
E90h	RA0PPS	_	_			RA0PF	PS<4:0>			00 0000	uu uuuu
E91h	RA1PPS	_	_			RA1PF	PS<4:0>			00 0000	uu uuuu
E92h	RA2PPS	—	_			RA2PF	PS<4:0>			00 0000	uu uuuu
E93h	RA3PPS	—	—			RA3PF	PS<4:0>			00 0000	uu uuuu
E94h	RA4PPS	_	_			RA4PF	PS<4:0>			00 0000	uu uuuu
E95h	RA5PPS	—	_			RA5PF	PS<4:0>			00 0000	uu uuuu
E96h	RA6PPS	_	_			RA6PF	PS<4:0>			00 0000	uu uuuu
E97h	RA7PPS	_	_			RA7PF	PS<4:0>			00 0000	uu uuuu
E98h	RB0PPS	—	_			RB0PF	PS<5:0>			00 0000	uu uuuu
E99h	RB1PPS	—	_			RB1PF	PS<5:0>			00 0000	uu uuuu
E9Ah	RB2PPS	—	—			RB2PF	PS<5:0>			00 0000	uu uuuu
E9Bh	RB3PPS	_	_			RB3PF	PS<5:0>			00 0000	uu uuuu
E9Ch	RB4PPS	—				RB4PF	PS<5:0>			00 0000	uu uuuu
E9Dh	RB5PPS	—	—			RB5PF	PS<5:0>			00 0000	uu uuuu
E9Eh	RB6PPS	—	—			RB6PF	PS<5:0>			00 0000	uu uuuu
E9Fh	RB7PPS	—	—			RB7PF	PS<5:0>			00 0000	uu uuuu
	RC0PPS	—	—				PS<5:0>			00 0000	uu uuuu
	RC1PPS	—	—				PS<5:0>			00 0000	uu uuuu
	RC2PPS						PS<5:0>			00 0000	uu uuuu
	RC3PPS						PS<5:0>			00 0000	uu uuuu
	RC4PPS	_	_				PS<5:0>			00 0000	uu uuuu
	RC5PPS	—	—				PS<5:0>			00 0000	uu uuuu
	RC6PPS	—	—				PS<5:0>			00 0000	uu uuuu
EA7h	RC7PPS	—	—			RC7PF	PS<5:0>			00 0000	uu uuuu

x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'. Legend:

Unimplemented, read as '1'. Note 1:

Unimplemented on PIC16LF1777/8/9. 2:

3: Unimplemented on PIC16(L)F1778.

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Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets	
Bank	29 (Cont.)											
EA8h	RD0PPS ⁽³⁾	_				RD0PF	'S<5:0>			00 0000	uu uuuu	
EA9h	RD1PPS ⁽³⁾	_	-			RD1PF	'S<5:0>			00 0000	uu uuuu	
EAAh	RD2PPS ⁽³⁾	_	-			RD2PF	'S<5:0>			00 0000	uu uuuu	
EABh	RD3PPS ⁽³⁾	_	_		RD3PPS<5:0> 00 0000 RD4PPS<5:0> 00 0000 RD5PPS<5:0> 00 0000							
EACh	RD4PPS ⁽³⁾	_	_		RD1PPS<5:0> 00 0000 - RD2PPS<5:0> 00 0000 - RD3PPS<5:0> 00 0000 - RD4PPS<5:0> 00 0000 - RD5PPS<5:0> 00 0000 - RD6PPS<5:0> 00 0000 - RD7PPS<5:0> 00 0000 - RD7PPS<5:0> 00 0000 - RD7PPS<5:0> 00 0000 - RD7PPS<5:0> 00 0000 -							
EADh	RD5PPS ⁽³⁾	_	_		RD1PPS<5:0> 00 0000 RD2PPS<5:0> 00 0000 RD3PPS<5:0> 00 0000 RD4PPS<5:0> 00 0000 RD5PPS<5:0> 00 0000 RD6PPS<5:0> 00 0000 RD7PPS<5:0> 00 0000 RD6PPS<5:0> 00 0000 RD7PPS<5:0> 00 0000 RD0 00 0000							
EAEh	RD6PPS ⁽³⁾	_	_		RD2PPS<5:0> 00 000 RD3PPS<5:0> 00 000 RD4PPS<5:0> 00 000 RD5PPS<5:0> 00 000 RD6PPS<5:0> 00 000 RD7PPS<5:0> 00 000 RD7PPS<5:0> 00 000 RD7PPS<5:0> 00 000 RD7PPS<5:0> 00 000							
EAFh	RD7PPS ⁽³⁾	_	_			RD7PF	'S<5:0>			00 0000	uu uuuu	
EB0h	RE0PPS ⁽³⁾	_	_			RE0PF	S<5:0>			00 0000	uu uuuu	
EB1h	RE1PPS ⁽³⁾	_	_			RE1PF	S<5:0>			00 0000	uu uuuu	
EB2h	RE2PPS ⁽³⁾	_	_				00 0000	uu uuuu				
EB3h to EEFh	_	Unimplemented								_	_	

|

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Ban	k 30										
F0Ch — F0Eh	_	Unimplemented								_	—
F0Fh	CLCDATA	—	—	_	_	MLC4OUT	MLC3OUT	MLC2OUT	MLC1OUT	x000	u000
F10h	CLC1CON	EN	—	OUT	INTP	INTN		MODE<2:0>		0-00 0000	0-00 0000
F11h	CLC1POL	POL	—	_		G4POL	G3POL	G2POL	G1POL	0 xxxx	0 uuuu
F12h	CLC1SEL0	—	—			D1S·	<5:0>			xx xxxx	uu uuuu
F13h	CLC1SEL1	—	_			D2S·	<5:0>			xx xxxx	uu uuuu
F14h	CLC1SEL2	—	_			D3S-	<5:0>			xx xxxx	uu uuuu
F15h	CLC1SEL3	—	_			D4S·	<5:0>			xx xxxx	uu uuuu
F16h	CLC1GLS0	G1D4T	G1D4N	G1D3T	G1D3N	G1D2T	G1D2N	G1D1T	G1D1N	xxxx xxxx	uuuu uuuu
F17h	CLC1GLS1	G2D4T	G2D4N	G2D3T	G2D3N	G2D2T	G2D2N	G2D1T	G2D1N	XXXX XXXX	uuuu uuuu
F18h	CLC1GLS2	G3D4T	G3D4N	G3D3T	G3D3N	G3D2T	G3D2N	G3D1T	G3D1N	XXXX XXXX	uuuu uuuu
F19h	CLC1GLS3	G4D4T	G4D4N	G4D3T	G4D3N	G4D2T	G4D2N	G4D1T	G4D1N	XXXX XXXX	uuuu uuuu
F1Ah	CLC2CON	EN	_	OUT	INTP	INTN		MODE<2:0>		0-00 0000	0-00 0000
F1Bh	CLC2POL	POL	_			G4POL	G3POL	G2POL	G1POL	0 xxxx	0 uuuu
F1Ch	CLC2SEL0	—	_			D1S·	<5:0>			xx xxxx	uu uuuu
F1Dh	CLC2SEL1	—	—			D2S-	<5:0>			xx xxxx	uu uuuu
F1Eh	CLC2SEL2	—	—			D3S-	<5:0>			xx xxxx	uu uuuu
F1Fh	CLC2SEL3	—	_	D4S<5:0>						xx xxxx	uu uuuu
F20h	CLC2GLS0	G1D4T	G1D4N	G1D3T	G1D3N	G1D2T	G1D2N	G1D1T	G1D1N	XXXX XXXX	uuuu uuuu
F21h	CLC2GLS1	G2D4T	G2D4N	G2D3T	G2D3N	G2D2T	G2D2N	G2D1T	G2D1N	XXXX XXXX	uuuu uuuu
F22h	CLC2GLS2	G3D4T	G3D4N	G3D3T	G3D3N	G3D2T	G3D2N	G3D1T	G3D1N	XXXX XXXX	uuuu uuuu
F23h	CLC2GLS3	G4D4T	G4D4N	G4D3T	G4D3N	G4D2T	G4D2N	G4D1T	G4D1N	XXXX XXXX	uuuu uuuu
F24h	CLC3CON	EN	—	OUT INTP INTN MODE<2:0>					0-00 0000	0-00 0000	
F25h	CLC3POL	POL	—	—	_	G4POL	G3POL	G2POL	G1POL	0 xxxx	0 uuuu
F26h	CLC3SEL0	—	—			D1S·	<5:0>			xx xxxx	uu uuuu
F27h	CLC3SEL1	—	—			D2S-	<5:0>			xx xxxx	uu uuuu
F28h	CLC3SEL2	_	—			D3S-	<5:0>			xx xxxx	uu uuuu
F29h	CLC3SEL3	—	_			D4S·	<5:0>			xx xxxx	uu uuuu
F2Ah	CLC3GLS0	G1D4T	G1D4N	G1D3T	G1D3N	G1D2T	G1D2N	G1D1T	G1D1N	xxxx xxxx	uuuu uuuu

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Bank	30 (Continued)										
F2Bh	CLC3GLS1	G2D4T	G2D4N	G2D3T	G2D3N	G2D2T	G2D2N	G2D1T	G2D1N	xxxx xxxx	uuuu uuuu
F2Ch	CLC3GLS2	G3D4T	G3D4N	G3D3T	G3D3N	G3D2T	G3D2N	G3D1T	G3D1N	xxxx xxxx	uuuu uuuu
F2Dh	CLC3GLS3	G4D4T	G4D4N	G4D3T	G4D3N	G4D2T	G4D2N	G4D1T	G4D1N	xxxx xxxx	uuuu uuuu
F2Eh	CLC4CON	EN	OE	OUT	INTP	INTN		MODE<2:0>			0000 0000
F2Fh	CLC4POL	POL	_	—	—	G4POL	G3POL	G2POL	G1POL	0 xxxx	0 uuuu
F30h	CLC4SEL0		D1S<7:0>								
F31h	CLC4SEL1		D2S<7:0>								uuuu uuuu
F32h	CLC4SEL2		D3S<7:0>								uuuu uuuu
F33h	CLC4SEL3		D4S<7:0>								uuuu uuuu
F34h	CLC4GLS0	G1D4T	G1D4N	G1D3T	G1D3N	G1D2T	G1D2N	G1D1T	G1D1N	xxxx xxxx	uuuu uuuu
F35h	CLC4GLS1	G2D4T	G2D4N	G2D3T	G2D3N	G2D2T	G2D2N	G2D1T	G2D1N	xxxx xxxx	uuuu uuuu
F36h	CLC4GLS2	G3D4T	G3D4N	G3D3T	G3D3N	G3D2T	G3D2N	G3D1T	G3D1N	xxxx xxxx	uuuu uuuu
F37h	CLC4GLS3	G4D4T	G4D4N	G4D3T	G4D3N	G4D2T	G4D2N	G4D1T	G4D1N	XXXX XXXX	uuuu uuuu
F2Eh — F6Fh	_	Unimplemented								-	_

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

					(000)						
Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Bank	Bank 31										
F8Ch to FE3h	_	Unimplemented									_
FE4h	STATUS_SHAD	—	—		—	_	Z	DC	С	xxx	uuu
FE5h	WREG_SHAD	Working Register	Shadow							XXXX XXXX	uuuu uuuu
FE6h	BSR_SHAD	—	—		Bank Select Regi	ster Shadow				x xxxx	u uuuu
FE7h	PCLATH_SHAD	—	Program Counter Latch High Register Shadow								
FE8h	FSR0L_SHAD	Indirect Data Men	Indirect Data Memory Address 0 Low Pointer Shadow								
FE9h	FSR0H_SHAD	Indirect Data Men	Indirect Data Memory Address 0 High Pointer Shadow								uuuu uuuu
FEAh	FSR1L_SHAD	Indirect Data Men	Indirect Data Memory Address 1 Low Pointer Shadow								uuuu uuuu
FEBh	FSR1H_SHAD	Indirect Data Memory Address 1 High Pointer Shadow								XXXX XXXX	uuuu uuuu
FECh	_	Unimplemented								—	—
FEDh	STKPTR	— — Current Stack Pointer								1 1111	1 1111
FEEh	TOSL	Top of Stack Low byte								XXXX XXXX	uuuu uuuu
FEFh	TOSH	_	Top of Stack High	byte						-xxx xxxx	-uuu uuuu

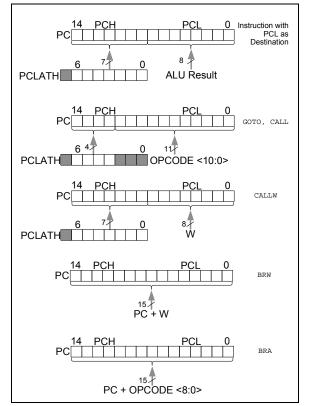
Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16LF1777/8/9.

3.5 PCL and PCLATH

The Program Counter (PC) is 15 bits wide. The low byte comes from the PCL register, which is a readable and writable register. The high byte (PC<14:8>) is not directly readable or writable and comes from PCLATH. On any Reset, the PC is cleared. Figure 3-4 shows the five situations for the loading of the PC.

FIGURE 3-4: LOADING OF PC IN DIFFERENT SITUATIONS



3.5.1 MODIFYING PCL

Executing any instruction with the PCL register as the destination simultaneously causes the Program Counter PC<14:8> bits (PCH) to be replaced by the contents of the PCLATH register. This allows the entire contents of the program counter to be changed by writing the desired upper seven bits to the PCLATH register. When the lower eight bits are written to the PCL register, all 15 bits of the program counter will change to the values contained in the PCLATH register.

3.5.2 COMPUTED GOTO

A computed GOTO is accomplished by adding an offset to the program counter (ADDWF PCL). When performing a table read using a computed GOTO method, care should be exercised if the table location crosses a PCL memory boundary (each 256-byte block). Refer to Application Note AN556, *"Implementing a Table Read"* (DS00556).

3.5.3 COMPUTED FUNCTION CALLS

A computed function CALL allows programs to maintain tables of functions and provide another way to execute state machines or look-up tables. When performing a table read using a computed function CALL, care should be exercised if the table location crosses a PCL memory boundary (each 256-byte block).

If using the CALL instruction, the PCH<2:0> and PCL registers are loaded with the operand of the CALL instruction. PCH<6:3> is loaded with PCLATH<6:3>.

The CALLW instruction enables computed calls by combining PCLATH and W to form the destination address. A computed CALLW is accomplished by loading the W register with the desired address and executing CALLW. The PCL register is loaded with the value of W and PCH is loaded with PCLATH.

3.5.4 BRANCHING

The branching instructions add an offset to the PC. This allows relocatable code and code that crosses page boundaries. There are two forms of branching, BRW and BRA. The PC will have incremented to fetch the next instruction in both cases. When using either branching instruction, a PCL memory boundary may be crossed.

If using BRW, load the W register with the desired unsigned address and execute BRW. The entire PC will be loaded with the address PC + 1 + W.

If using BRA, the entire PC will be loaded with PC + 1 +, the signed value of the operand of the BRA instruction.

3.6 Stack

All devices have a 16-level x 15-bit wide hardware stack (refer to Figures 3-1 and 3-2). The stack space is not part of either program or data space. The PC is PUSHed onto the stack when CALL or CALLW instructions are executed or an interrupt causes a branch. The stack is POPed in the event of a RETURN, RETLW or a RETFIE instruction execution. PCLATH is not affected by a PUSH or POP operation.

The stack operates as a circular buffer if the STVREN bit is programmed to '0' (Configuration Words). This means that after the stack has been PUSHed sixteen times, the seventeenth PUSH overwrites the value that was stored from the first PUSH. The eighteenth PUSH overwrites the second PUSH (and so on). The STKOVF and STKUNF flag bits will be set on an Overflow/Underflow, regardless of whether the Reset is enabled.

Note: There are no instructions/mnemonics called PUSH or POP. These are actions that occur from the execution of the CALL, CALLW, RETURN, RETLW and RETFIE instructions or the vectoring to an interrupt address.

3.6.1 ACCESSING THE STACK

The stack is available through the TOSH, TOSL and STKPTR registers. STKPTR is the current value of the Stack Pointer. The TOSH:TOSL register pair points to the TOP of the stack. Both registers are read/writable. TOS is split into TOSH and TOSL due to the 15-bit size of the PC. To access the stack, adjust the value of STKPTR, which will position TOSH:TOSL, then read/write to TOSH:TOSL. STKPTR is five bits to allow detection of overflow and underflow.

Note:	Care should be taken when modifying the
	STKPTR while interrupts are enabled.

During normal program operation, CALL, CALLW and Interrupts will increment STKPTR while RETLW, RETURN, and RETFIE will decrement STKPTR. At any time, STKPTR can be inspected to see how much stack is left. The STKPTR always points at the currently used place on the stack. Therefore, a CALL or CALLW will increment the STKPTR and then write the PC, and a return will unload the PC and then decrement the STKPTR.

Reference Figure 3-5 through Figure 3-8 for examples of accessing the stack.

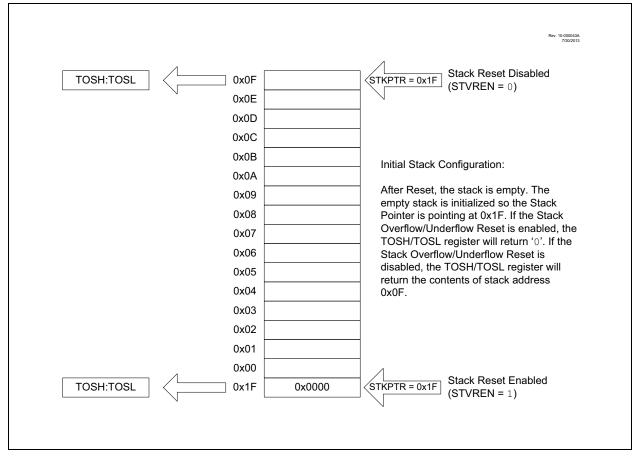
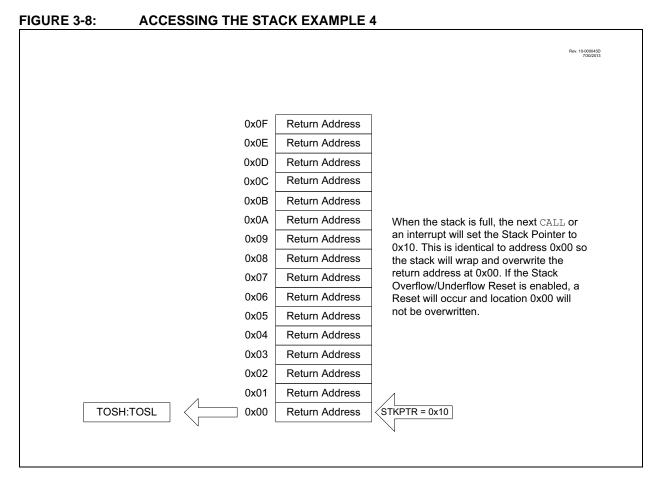


FIGURE 3-5: ACCESSING THE STACK EXAMPLE 1

0x0F				
OXOF				Rev. 10-0000438 7/30/2013
OXOF				
0x00		0x0)F	
Image: state of the state		0xC)E	
OXOF This figure shows the stack configuration of the first CALL or a single interrupt. If a REPUTINN instruction is executed, the program Counter and the Stack Pointer decremented to the empty state (0x1F). TOSH:TOSL OXOF NOTE OXOF OXOF Return Address OXOF OXOF		0x0	D	
with a stack configuration of the the first CALL or a single interrupt, the stack configuration after the first CALL or a single interrupt, the stack configuration of the the most of the program Counter and the Stack Chiner decremented to the empty state (0x1F). with a star to		0x0	OC	
0x09		0xC)B	
0x08		0x0	A	
0x00 0x07 0x06 If a RETURN instruction is executed, the return address will be placed in the Program Counter and the Stack Pointer decremented to the empty state (0x1F). 0x04 0x04 0x04 0x04 0x04 0x04 0x04 0x04		0x0)9	
0x06 Program Counter and the Stack Pointer decremented to the empty state (0x1F). 0x03 0x04 0x03 0x02 0x01 Return Address TOSH:TOSL 0x00 Return Address TKPTR = 0x00				If a RETURN instruction is executed, the
0x005				
0x04 0x03 0x02 0x01 0x00 Image: Construction of the second se				
OX03 OX02 OX01 Return Address ST.PTR = 0x00 GURE 3-7: ACCESSING THE STACK EXAMPLE 3 ST.PTR = 0x00 Mathematical Stress Mathematical Stress Mathematical Stress (Content of the stress of Return Address of Return Address of Return addresses into the Program Counter and pop the stack.				
OX02 OX01 Return Address ST/PTR = 0x00 GURE 3-7: ACCESSING THE STACK EXAMPLE 3 STORE STACK EXAMPLE 3 OXOF OXOF OXOF OXOE OXOE OXOB OXOE <td></td> <td></td> <td></td> <td></td>				
TOSH:TOSL 0x01 Return Address STAPTR = 0x00 GURE 3-7: ACCESSING THE STACK EXAMPLE 3 Cover and the stack of the stack looks like the figure on the left. A series of RETURN instructions will repeatedly place the return addresses into the Program Counter and pop the stack. TOSH:TOSL 0x07 Return Address				
TOSH:TOSL 0x00 Return Address TKPTR = 0x00 GURE 3.7: ACCESSING THE STACK EXAMPLE 3 Control of the stack lock size of the stack lock size of the stack lock size of the figure on the left. A series of RETURN instructions will repeatedly place the return addresses into the Program Counter and pop the stack. TOSH:TOSL 0x06 Return Address STKPTR = 0x06				
Instruction ACCESSING THE STACK EXAMPLE 3 Instruction Instruction Instruction <td< td=""><td>TOSE</td><td> /</td><td></td><td></td></td<>	TOSE	/		
0x0F				
0x0E	IGURE 3-7:	ACCESSING THE STA	CK EXAMPLE 3	3
0x0E	GURE 3-7:	ACCESSING THE STA	CK EXAMPLE 3	Rev 10-00043C
0x0D	GURE 3-7:	ACCESSING THE STA	CK EXAMPLE 3	Rev 10-00043C
0x0C After seven CALLS or six CALLS and an interrupt, the stack looks like the figure on the left. A series of RETURN instructions will repeatedly place the return addresses into the Program Counter and pop the stack. 0x08 0x07 0x07 STKPTR = 0x06	GURE 3-7:		CK EXAMPLE 3	Rev. 10-00043C
0x0B interseven CALLS of SIX CALLS and an interrupt, the stack looks like the figure on the left. A series of RETURN instructions will repeatedly place the return addresses into the Program Counter and pop the stack. 0x0A 0x0A 0x0A 0x0A 0x0B 0x0A 0x0A 0x0A 0x0B 0x0A 0x0A 0x0B 0x0B 0x0B 0x0B 0x0A 0x0A 0x0B 0x0B 0x0A 0x0A 0x0B 0x0B 0x0B 0x0C Return Address STKPTR = 0x06 STKPTR = 0x06	GURE 3-7:	0x0F	CK EXAMPLE 3	Rev. 10-00043C
0x0A the left. A series of RETURN instructions will repeatedly place the return addresses into the Program Counter and pop the stack. 0x08 0x07 TOSH:TOSL 0x06 0x06 Return Address STKPTR = 0x06	GURE 3-7:	0x0F 0x0E 0x0D	CK EXAMPLE 3	Rev. 10-00043C
0x0A repeatedly place the return addresses into the Program Counter and pop the stack. 0x08 0x07 TOSH:TOSL 0x06 Return Address STKPTR = 0x06	GURE 3-7:	0x0F 0x0E 0x0D 0x0C	CK EXAMPLE 3	Rev. 10-000033C 77902013
0x08 0x07 TOSH:TOSL 0x06 Return Address	GURE 3-7:	0x0F 0x0E 0x0D 0x0C 0x0C 0x0B	CK EXAMPLE 3	After seven CALLs or six CALLs and an interrupt, the stack looks like the figure on
0x07 TOSH:TOSL 0x06 Return Address STKPTR = 0x06	GURE 3-7:	0x0F 0x0E 0x0D 0x0C 0x0B 0x0A	CK EXAMPLE 3	After seven CALLS or six CALLS and an interrupt, the stack looks like the figure on the left. A series of RETURN instructions will repeatedly place the return addresses into
TOSH:TOSL 0x06 Return Address STKPTR = 0x06	GURE 3-7:	0x0F 0x0E 0x0D 0x0C 0x0B 0x0B 0x0A 0x09		After seven CALLS or six CALLS and an interrupt, the stack looks like the figure on the left. A series of RETURN instructions will repeatedly place the return addresses into
	GURE 3-7:	0x0F 0x0E 0x0D 0x0C 0x0B 0x0A 0x09 0x08	CK EXAMPLE 3	After seven CALLS or six CALLS and an interrupt, the stack looks like the figure on the left. A series of RETURN instructions will repeatedly place the return addresses into
		0x0F 0x0E 0x0D 0x0C 0x0B 0x0A 0x0A 0x09 0x08 0x07		After seven CALLS or six CALLS and an interrupt, the stack looks like the figure on the left. A series of RETURN instructions will repeatedly place the return addresses into the Program Counter and pop the stack.
		0x0F 0x0E 0x0D 0x0C 0x0B 0x0A 0x09 0x08 0x07 TTOSL 0x06	Return Address	After seven CALLS or six CALLS and an interrupt, the stack looks like the figure on the left. A series of RETURN instructions will repeatedly place the return addresses into the Program Counter and pop the stack.
		0x0F 0x0E 0x0D 0x0C 0x0B 0x0A 0x09 0x08 0x07 0x06 0x05	Return Address Return Address	After seven CALLS or six CALLS and an interrupt, the stack looks like the figure on the left. A series of RETURN instructions will repeatedly place the return addresses into the Program Counter and pop the stack.
		0x0F 0x0E 0x0D 0x0C 0x0B 0x0A 0x09 0x08 0x09 0x08 0x07 0x06 0x05 0x04	Return Address Return Address Return Address	After seven CALLS or six CALLS and an interrupt, the stack looks like the figure on the left. A series of RETURN instructions will repeatedly place the return addresses into the Program Counter and pop the stack.
0x02 Return Address		0x0F 0x0E 0x0D 0x0C 0x0B 0x0A 0x08 0x07 0x08 0x07 0x06 0x05 0x04 0x03	Return Address Return Address Return Address Return Address Return Address	After seven CALLS or six CALLS and an interrupt, the stack looks like the figure on the left. A series of RETURN instructions will repeatedly place the return addresses into the Program Counter and pop the stack.
0x02 Return Address 0x01 Return Address		0x0F 0x0E 0x0D 0x0C 0x0B 0x0A 0x08 0x03 0x06 0x05 0x04 0x03 0x02	Return Address Return Address Return Address Return Address Return Address Return Address	After seven CALLS or six CALLS and an interrupt, the stack looks like the figure on the left. A series of RETURN instructions will repeatedly place the return addresses into the Program Counter and pop the stack.
0x02 Return Address		0x0F 0x0E 0x0D 0x0C 0x0B 0x0A 0x08 0x07 0x08 0x07 0x06 0x05 0x04 0x03	Return Address Return Address Return Address Return Address Return Address	After seven CALLS or six CALLS and an interrupt, the stack looks like the figure on the left. A series of RETURN instructions will repeatedly place the return addresses into the Program Counter and pop the stack.
		0x0F 0x0E 0x0D 0x0C 0x0B 0x0A 0x08 0x03 0x06 0x05 0x04 0x03 0x02	Return Address Return Address Return Address Return Address Return Address Return Address	After seven CALLS or six CALLS and an interrupt, the stack looks like the figure on the left. A series of RETURN instructions will repeatedly place the return addresses into the Program Counter and pop the stack.



3.6.2 OVERFLOW/UNDERFLOW RESET

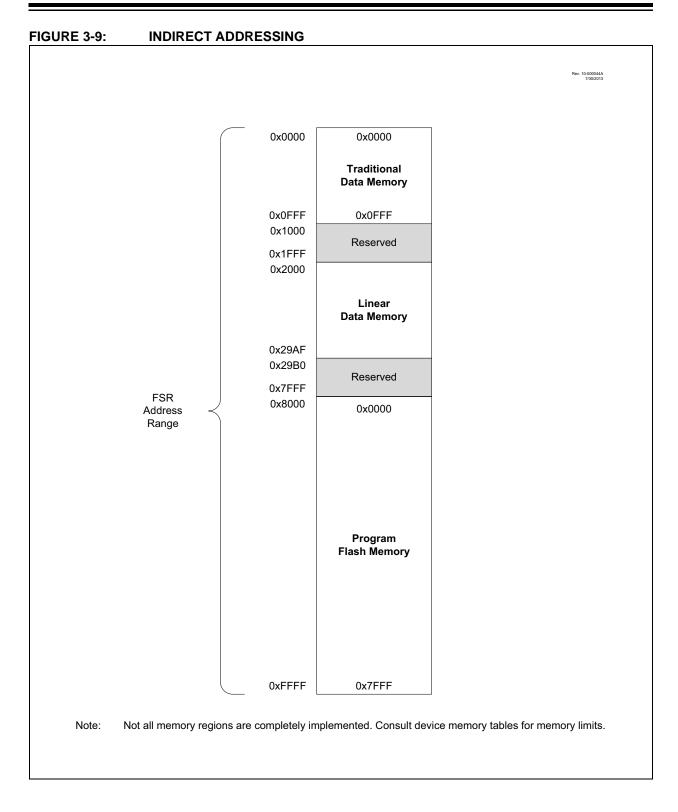
If the STVREN bit in Configuration Words is programmed to '1', the device will be reset if the stack is PUSHed beyond the sixteenth level or POPed beyond the first level, setting the appropriate bits (STKOVF or STKUNF, respectively) in the PCON register.

3.7 Indirect Addressing

The INDFn registers are not physical registers. Any instruction that accesses an INDFn register actually accesses the register at the address specified by the File Select Registers (FSR). If the FSRn address specifies one of the two INDFn registers, the read will return '0' and the write will not occur (though Status bits may be affected). The FSRn register value is created by the pair FSRnH and FSRnL.

The FSR registers form a 16-bit address that allows an addressing space with 65536 locations. These locations are divided into three memory regions:

- · Traditional Data Memory
- · Linear Data Memory
- Program Flash Memory



3.7.1 TRADITIONAL DATA MEMORY

The traditional data memory is a region from FSR address 0x000 to FSR address 0xFFF. The addresses correspond to the absolute addresses of all SFR, GPR and common registers.

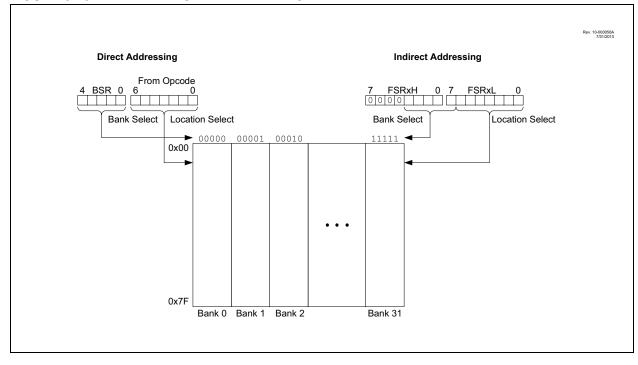


FIGURE 3-10: TRADITIONAL DATA MEMORY MAP

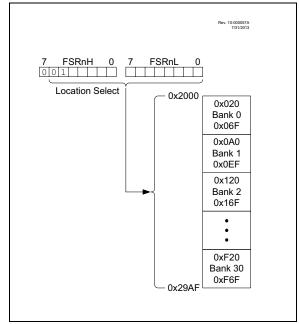
3.7.2 LINEAR DATA MEMORY

The linear data memory is the region from FSR address 0x2000 to FSR address 0x29AF. This region is a virtual region that points back to the 80-byte blocks of GPR memory in all the banks.

Unimplemented memory reads as 0x00. Use of the linear data memory region allows buffers to be larger than 80 bytes because incrementing the FSR beyond one bank will go directly to the GPR memory of the next bank.

The 16 bytes of common memory are not included in the linear data memory region.

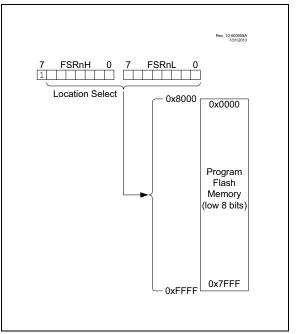
FIGURE 3-11: LINEAR DATA MEMORY MAP



3.7.3 PROGRAM FLASH MEMORY

To make constant data access easier, the entire program Flash memory is mapped to the upper half of the FSR address space. When the MSB of FSRnH is set, the lower 15 bits are the address in program memory which will be accessed through INDF. Only the lower eight bits of each memory location is accessible via INDF. Writing to the program Flash memory cannot be accomplished via the FSR/INDF interface. All instructions that access program Flash memory via the FSR/INDF interface will require one additional instruction cycle to complete.

FIGURE 3-12: PROGRAM FLASH MEMORY MAP



4.0 DEVICE CONFIGURATION

Device configuration consists of Configuration Words, Code Protection and Device ID.

4.1 Configuration Words

There are several Configuration Word bits that allow different oscillator and memory protection options. These are implemented as Configuration Word 1 at 8007h and Configuration Word 2 at 8008h.

Note: The DEBUG bit in Configuration Words is managed automatically by device development tools including debuggers and programmers. For normal device operation, this bit should be maintained as a '1'.

4.2 Register Definitions: Configuration Words

R/P-1 R/P-1 R/P-1 R/P-1 R/P-1 U-1 **FCMEN IESO** CLKOUTEN BOREN<1:0> ____ bit 8 bit 13 R/P-1 R/P-1 R/P-1 R/P-1 **R/P-1** R/P-1 R/P-1 R/P-1 CP(1) PWRTE MCLRE WDTE<1:0> FOSC<2:0> bit 7 bit 0 Legend: R = Readable bit P = Programmable bit U = Unimplemented bit, read as '1' '0' = Bit is cleared '1' = Bit is set -n = Value when blank or after Bulk Erase bit 13 FCMEN: Fail-Safe Clock Monitor Enable bit 1 = ONFail-Safe Clock Monitor and internal/external switchover are both enabled 0 = OFFFail-Safe Clock Monitor is disabled IESO: Internal External Switchover bit bit 12 1 = ONInternal/External Switchover mode is enabled 0 = OFFInternal/External Switchover mode is disabled **CLKOUTEN:** Clock Out Enable bit bit 11 If FOSC Configuration bits are set to LP, XT, HS modes: This bit is ignored, CLKOUT function is disabled. Oscillator function on the CLKOUT pin. All other FOSC modes: 1 = ONCLKOUT function is disabled. I/O function on the CLKOUT pin. 0 = OFFCLKOUT function is enabled on the CLKOUT pin. bit 10-9 BOREN<1:0>: Brown-out Reset Enable bits 11 = ONBOR enabled 10 = NSLEEP BOR enabled during operation and disabled in Sleep 01 = SBODEN BOR controlled by SBOREN bit of the BORCON register BOR disabled 00 = OFFbit 8 Unimplemented: Read as '1' bit 7 **CP:** Code Protection bit⁽¹⁾ 1 = OFFProgram memory code protection is disabled Program memory code protection is enabled 0 = ONMCLRE: MCLR/VPP Pin Function Select bit bit 6 If LVP bit = 1: This bit is ignored. If LVP bit = 0: 1 = ONMCLR/VPP pin function is MCLR; Weak pull-up enabled. 0 = OFFMCLR/VPP pin function is digital input; MCLR internally disabled; Weak pull-up under control of WPUA3 bit. bit 5 **PWRTE:** Power-up Timer Enable bit 1 = OFFPWRT disabled PWRT enabled 0 = **ON** WDTE<1:0>: Watchdog Timer Enable bit bit 4-3 WDT enabled 11 = ON10 = NSLEEP WDT enabled while running and disabled in Sleep 01 = SWDTEN WDT controlled by the SWDTEN bit in the WDTCON register 00 = OFFWDT disabled

REGISTER 4-1: CONFIG1: CONFIGURATION WORD 1

REGISTER 4-1: CONFIG1: CONFIGURATION WORD 1 (CONTINUED)

bit 2-0	FOSC<2:0>: Oscillator Selection bits
---------	--------------------------------------

- 111 = ECH External Clock, High-Power mode: CLKIN supplied to OSC1/CLKIN pin
- 110 = ECM External Clock, Medium Power mode: CLKIN supplied to OSC1/CLKIN pin
- 101 = ECL External Clock, Low-Power mode: CLKIN supplied to OSC1/CLKIN pin
- 100 = INTOSC Internal HFINTOSC. I/O function on CLKIN pin
- 011 = EXTRC External RC circuit connected to CLKIN pin
- 010 = HS High-speed crystal/resonator connected between OSC1 and OSC2 pins
- 001 = XT Crystal/resonator connected between OSC1 and OSC2 pins
- 000 = LP Low-power crystal connected between OSC1 and OSC2 pins
- **Note 1:** The entire Flash program memory will be erased when the code protection is turned off during an erase. When a Bulk Erase Program Memory command is executed, the entire program Flash memory and configuration memory will be erased.

		R/P-1	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1
		LVP ⁽¹⁾	DEBUG ⁽²⁾	LPBOR	BORV ⁽³⁾	STVREN	PLLEN
		bit 13					bit 8
R/P-1	U-1	U-1	U-1	U-1	R/P-1	R/P-1	R/P-1
ZCD	—	—	—	—	PPS1WAY	WRT<	<1:0>
bit 7							bit
Legend:	a hit		abla bit		nted hit read as	(4)	
R = Readable		P = Programma	adie dit	•	ented bit, read as		
'0' = Bit is cle	ared	'1' = Bit is set		-n = value whe	n blank or after B	uik Erase	
bit 13	LVP: Low-Volt 1 = ON 0 = OFF	tage Programming Low-voltage prog High-voltage on	gramming enable		ming		
bit 12	DEBUG : In-Ci 1 = OFF 0 = ON	-	ger disabled, ICS		-	purpose I/O pins ed to the debugge	
bit 11	LPBOR: Low- 1 = OFF 0 = ON	Power BOR Enat Low-Power Brov Low-Power Brov	vn-out Reset is d				
bit 10	BORV: Brown 1 = LO 0 = HI	-out Reset Voltag Brown-out Rese Brown-out Rese	t voltage (VBOR),				
bit 9	STVREN: Sta 1 = ON 0 = OFF	ck Overflow/Unde Stack Overflow o Stack Overflow o	or Underflow will	cause a Reset	et		
bit 8	PLLEN: PLL E 1 = ON 0 = OFF	Enable bit 4xPLL enabled 4xPLL disabled					
bit 7	ZCD : ZCD En 1 = OFF 0 = ON			ed by setting the	ZCDSEN bit of 2	ZCDCON	
bit 6-3	Unimplement	ed: Read as '1'					
bit 2	PPS1WAY : PF 1 = ON 0 = OFF	is set, all future of	bit can only be se changes to PPS	et once after an u registers are pre	vented	ce is executed; o unlocking sequen	
bit 1-0	$\frac{4 \text{ kW Flash m}}{11 = 0}$ $10 = B$ $01 = H$ $00 = A$	OOT 0000h IALF 0000h ILL 0000h emory (PIC16(L)F DFF Write GOOT 0000h IALF 0000h	<u>1764/8)</u> protection off to 01FFh write p to 07FFh write p to 0FFFh write p <u>1765/9)</u> protection off to 01FFh write p to 0FFFh write p	protected, 0200h protected, 0800h protected, no ado protected, 0200h protected, 1000h	to 0FFFh may be dresses may be n to 1FFFh may be to 1FFFh may be	e modified by PM e modified by PM nodified by PMCC e modified by PM e modified by PM nodified by PMCC	CON control DN control CON control CON control
2: 1	00 = A The <u>LVP bit</u> canno The DEBUG bit in	LL 0000h t be programmed Configuration Wo	to 1FFFh write p to '0' when Prog rds is managed a	protected, no add ramming mode i automatically by	dresses may be n s entered via LVF	nodified by PMCC p. ent tools includin	ON control

REGISTER 4-2: CONFIG2: CONFIGURATION WORD 2

and programmers. For normal device operation, this bit should be maintained as a '1'.

3: See VBOR parameter for specific trip point voltages.

4.3 Code Protection

Code protection allows the device to be protected from unauthorized access. Program memory protection is controlled independently. Internal access to the program memory is unaffected by any code protection setting.

4.3.1 PROGRAM MEMORY PROTECTION

The entire program memory space is protected from external reads and writes by the \overline{CP} bit in Configuration Words. When $\overline{CP} = 0$, external reads and writes of program memory are inhibited and a read will return all '0's. The CPU can continue to read program memory, regardless of the protection bit settings. Writing the program memory is dependent upon the write protection setting. See Section 4.4 "Write Protection" for more information.

4.4 Write Protection

Write protection allows the device to be protected from unintended self-writes. Applications, such as boot loader software, can be protected while allowing other regions of the program memory to be modified.

The WRT<1:0> bits in Configuration Words define the size of the program memory block that is protected.

4.5 User ID

Four memory locations (8000h-8003h) are designated as ID locations where the user can store checksum or other code identification numbers. These locations are readable and writable during normal execution. See **Section 10.4 "User ID, Device ID and Configuration Word Access"** for more information on accessing these memory locations. For more information on checksum calculation, see the *"PIC16(L)F170X Memory Programming Specification"* (DS41683).

4.6 Device ID and Revision ID

The 14-bit device ID word is located at 8006h and the 14-bit revision ID is located at 8005h. These locations are read-only and cannot be erased or modified. See **Section 10.4 "User ID, Device ID and Configuration Word Access"** for more information on accessing these memory locations.

Development tools, such as device programmers and debuggers, may be used to read the Device ID and Revision ID.

4.7 Register Definitions: Device and Revision

REGISTER 4-3: DEVID: DEVICE ID REGISTER

	R	R	R	R	R	R					
			DEV<	:13:8>							
	bit 13	bit 13 bit 8									
R	R	R	R	R	R	R					
DEV<7:0>											
						bit 0					
	R		bit 13 R R R	DEV bit 13	R R R R R R	DEV<13:8> bit 13					

Legend:

R = Readable bit

'1' = Bit is set '0' = Bit is cleared

bit 13-0 **DEV<13:0>:** Device ID bits

Device	DEVID<13:0> Values
PIC16F1777	11 0000 1000 1110 (308E)
PIC16F1778	11 0000 1000 1111 (308F)
PIC16F1779	11 0000 1001 0000 (3090)
PIC16LF1777	11 0000 1001 0001 (3091)
PIC16LF1778	11 0000 1001 0010 (3092)
PIC16LF1779	11 0000 1001 0011 (3093)

REGISTER 4-4: REVID: REVISION ID REGISTER

		-					
		R	R	R	R	R	R
				REV<	:13:8>		
		bit 13					bit 8
R	R	R	R	R	R	R	R
			REV	<7:0>			
bit 7							bit 0
Legend:							
R = Readable bit	t						
'1' = Bit is set		'0' = Bit is cleare	ed				

bit 13-0 **REV<13:0>:** Revision ID bits

5.0 OSCILLATOR MODULE (WITH FAIL-SAFE CLOCK MONITOR)

5.1 Overview

The oscillator module has a wide variety of clock sources and selection features that allow it to be used in a wide range of applications while maximizing performance and minimizing power consumption. Figure 5-1 illustrates a block diagram of the oscillator module.

Clock sources can be supplied from external oscillators, quartz crystal resonators, ceramic resonators and Resistor-Capacitor (RC) circuits. In addition, the system clock source can be supplied from one of two internal oscillators and PLL circuits, with a choice of speeds selectable via software. Additional clock features include:

- Selectable system clock source between external or internal sources via software.
- Two-Speed Start-up mode, which minimizes latency between external oscillator start-up and code execution.
- Fail-Safe Clock Monitor (FSCM) designed to detect a failure of the external clock source (LP, XT, HS, ECH, ECM, ECL or EXTRC modes) and switch automatically to the internal oscillator.
- Oscillator Start-up Timer (OST) ensures stability of crystal oscillator sources.

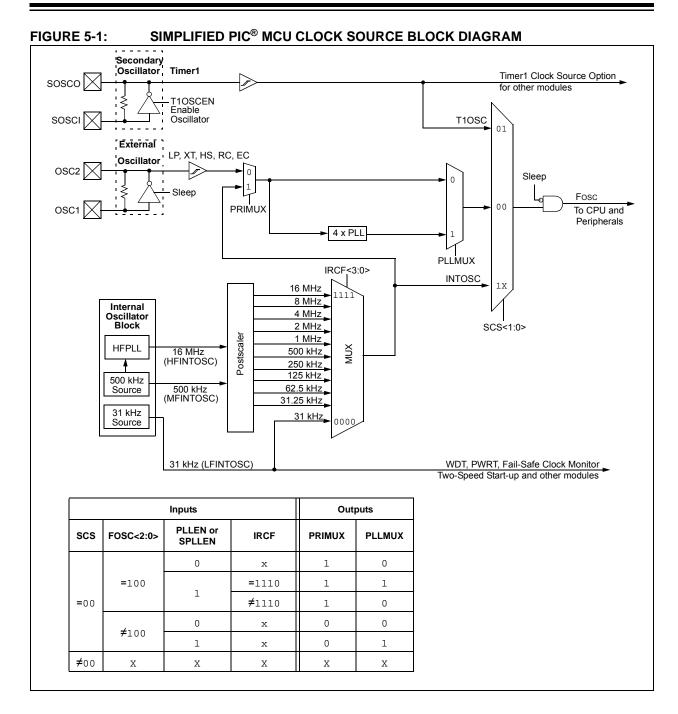
The oscillator module can be configured in one of the following clock modes.

- 1. ECL External Clock Low-Power mode (0 MHz to 0.5 MHz)
- 2. ECM External Clock Medium Power mode (0.5 MHz to 4 MHz)
- 3. ECH External Clock High-Power mode (4 MHz to 32 MHz)
- 4. LP 32 kHz Low-Power Crystal mode.
- 5. XT Medium Gain Crystal or Ceramic Resonator Oscillator mode (up to 4 MHz)
- 6. HS High Gain Crystal or Ceramic Resonator mode (4 MHz to 20 MHz)
- 7. EXTRC External Resistor-Capacitor
- 8. INTOSC Internal oscillator (31 kHz to 32 MHz)

Clock Source modes are selected by the FOSC<2:0> bits in the Configuration Words. The FOSC bits determine the type of oscillator that will be used when the device is first powered.

The ECH, ECM, and ECL clock modes rely on an external logic level signal as the device clock source. The LP, XT, and HS clock modes require an external crystal or resonator to be connected to the device. Each mode is optimized for a different frequency range. The EXTRC clock mode requires an external resistor and capacitor to set the oscillator frequency.

The INTOSC internal oscillator block produces low, medium, and high-frequency clock sources, designated LFINTOSC, MFINTOSC and HFINTOSC. (see Internal Oscillator Block, Figure 5-1). A wide selection of device clock frequencies may be derived from these three clock sources.



5.2 Clock Source Types

Clock sources can be classified as external or internal.

External clock sources rely on external circuitry for the clock source to function. Examples are: oscillator modules (ECH, ECM, ECL mode), quartz crystal resonators or ceramic resonators (LP, XT and HS modes) and Resistor-Capacitor (EXTRC) mode circuits.

Internal clock sources are contained within the oscillator module. The internal oscillator block has two internal oscillators and a dedicated Phase Lock Loop (HFPLL) that are used to generate three internal system clock sources: the 16 MHz High-Frequency Internal Oscillator (HFINTOSC), 500 kHz (MFINTOSC) and the 31 kHz Low-Frequency Internal Oscillator (LFINTOSC).

The system clock can be selected between external or internal clock sources via the System Clock Select (SCS) bits in the OSCCON register. See Section 5.3 "Clock Switching" for additional information.

5.2.1 EXTERNAL CLOCK SOURCES

An external clock source can be used as the device system clock by performing one of the following actions:

- Program the FOSC<2:0> bits in the Configuration Words to select an external clock source that will be used as the default system clock upon a device Reset.
- Write the SCS<1:0> bits in the OSCCON register to switch the system clock source to:
 - Secondary oscillator during run-time, or
 - An external clock source determined by the value of the FOSC bits.

See Section 5.3 "Clock Switching" for more information.

5.2.1.1 EC Mode

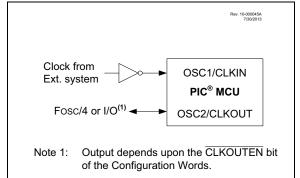
The External Clock (EC) mode allows an externally generated logic level signal to be the system clock source. When operating in this mode, an external clock source is connected to the OSC1 input. OSC2/CLKOUT is available for general purpose I/O or CLKOUT. Figure 5-2 shows the pin connections for EC mode.

EC mode has three power modes to select from through Configuration Words:

- ECH High power, 4-32 MHz
- ECM Medium power, 0.5-4 MHz
- ECL Low power, 0-0.5 MHz

The Oscillator Start-up Timer (OST) is disabled when EC mode is selected. Therefore, there is no delay in operation after a Power-on Reset (POR) or wake-up from Sleep. Because the PIC[®] MCU design is fully static, stopping the external clock input will have the effect of halting the device while leaving all data intact. Upon restarting the external clock, the device will resume operation as if no time had elapsed.

FIGURE 5-2: EXTERNAL CLOCK (EC) MODE OPERATION



5.2.1.2 LP, XT, HS Modes

The LP, XT and HS modes support the use of quartz crystal resonators or ceramic resonators connected to OSC1 and OSC2 (Figure 5-3). The three modes select a low, medium or high gain setting of the internal inverter-amplifier to support various resonator types and speed.

LP Oscillator mode selects the lowest gain setting of the internal inverter-amplifier. LP mode current consumption is the least of the three modes. This mode is designed to drive only 32.768 kHz tuning-fork type crystals (watch crystals).

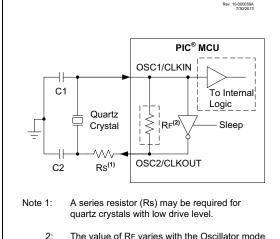
XT Oscillator mode selects the intermediate gain setting of the internal inverter-amplifier. XT mode current consumption is the medium of the three modes. This mode is best suited to drive resonators with a medium drive level specification.

HS Oscillator mode selects the highest gain setting of the internal inverter-amplifier. HS mode current consumption is the highest of the three modes. This mode is best suited for resonators that require a high drive setting.

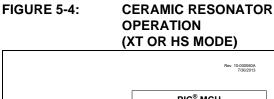
Figure 5-3 and Figure 5-4 show typical circuits for quartz crystal and ceramic resonators, respectively.

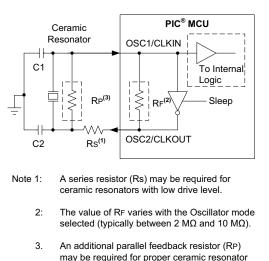
HS MODE)

FIGURE 5-3: QUARTZ CRYSTAL OPERATION (LP, XT OR



- 2. The value of RF values with the Oscillator mode selected (typically between 2 M Ω and 10 M Ω).
- Note 1: Quartz crystal characteristics vary according to type, package and manufacturer. The user should consult the manufacturer data sheets for specifications and recommended application.
 - 2: Always verify oscillator performance over the VDD and temperature range that is expected for the application.
 - **3:** For oscillator design assistance, reference the following Microchip Application Notes:
 - AN826, "Crystal Oscillator Basics and Crystal Selection for rfPIC[®] and PIC[®] Devices" (DS00826)
 - AN849, "Basic PIC[®] Oscillator Design" (DS00849)
 - AN943, "Practical PIC[®] Oscillator Analysis and Design" (DS00943)
 - AN949, "*Making Your Oscillator Work*" (DS00949)





5.2.1.3 Oscillator Start-up Timer (OST)

operation.

If the oscillator module is configured for LP, XT or HS modes, the Oscillator Start-up Timer (OST) counts 1024 oscillations from OSC1. This occurs following a Power-on Reset (POR) and when the Power-up Timer (PWRT) has expired (if configured), or a wake-up from Sleep. During this time, the program counter does not increment and program execution is suspended, unless either FSCM or Two-Speed Start-Up are enabled. In this case, code will continue to execute at the selected INTOSC frequency while the OST is counting. The OST ensures that the oscillator circuit, using a quartz crystal resonator or ceramic resonator, has started and is providing a stable system clock to the oscillator module.

In order to minimize latency between external oscillator start-up and code execution, the Two-Speed Clock Start-up mode can be selected (see Section 5.4 "Two-Speed Clock Start-up Mode").

5.2.1.4 4x PLL

The oscillator module contains a 4x PLL that can be used with both external and internal clock sources to provide a system clock source. The input frequency for the 4x PLL must fall within specifications. See the PLL Clock Timing Specifications in Table 36-9.

The 4x PLL may be enabled for use by one of two methods:

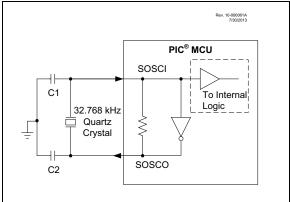
- 1. Program the PLLEN bit in Configuration Words to a '1'.
- Write the SPLLEN bit in the OSCCON register to a '1'. If the PLLEN bit in Configuration Words is programmed to a '1', then the value of SPLLEN is ignored.

5.2.1.5 Secondary Oscillator

The secondary oscillator is a separate crystal oscillator that is associated with the Timer1 peripheral. It is optimized for timekeeping operations with a 32.768 kHz crystal connected between the SOSCO and SOSCI device pins.

The secondary oscillator can be used as an alternate system clock source and can be selected during run-time using clock switching. Refer to **Section 5.3 "Clock Switching"** for more information.

FIGURE 5-5: QUARTZ CRYSTAL OPERATION (SECONDARY OSCILLATOR)



- Note 1: Quartz crystal characteristics vary according to type, package and manufacturer. The user should consult the manufacturer data sheets for specifications and recommended application.
 - 2: Always verify oscillator performance over the VDD and temperature range that is expected for the application.
 - **3:** For oscillator design assistance, reference the following Microchip Application Notes:
 - AN826, "Crystal Oscillator Basics and Crystal Selection for rfPIC[®] and PIC[®] Devices" (DS00826)
 - AN849, "Basic PIC[®] Oscillator Design" (DS00849)
 - AN943, "Practical PIC[®] Oscillator Analysis and Design" (DS00943)
 - AN949, "Making Your Oscillator Work" (DS00949)
 - TB097, "Interfacing a Micro Crystal MS1V-T1K 32.768 kHz Tuning Fork Crystal to a PIC16F690/SS" (DS91097)
 - AN1288, "Design Practices for Low-Power External Oscillators" (DS01288)

5.2.1.6 External RC Mode

The external Resistor-Capacitor (EXTRC) mode supports the use of an external RC circuit. This allows the designer maximum flexibility in frequency choice while keeping costs to a minimum when clock accuracy is not required.

The RC circuit connects to OSC1. OSC2/CLKOUT is available for general purpose I/O or CLKOUT. The function of the OSC2/CLKOUT pin is determined by the CLKOUTEN bit in Configuration Words.

Figure 5-6 shows the external RC mode connections.

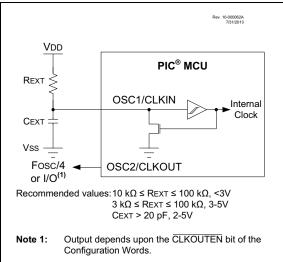


FIGURE 5-6: EXTERNAL RC MODES

The RC oscillator frequency is a function of the supply voltage, the resistor (REXT) and capacitor (CEXT) values and the operating temperature. Other factors affecting the oscillator frequency are:

- threshold voltage variation
- · component tolerances
- packaging variations in capacitance

The user also needs to take into account variation due to tolerance of external RC components used.

5.2.2 INTERNAL CLOCK SOURCES

The device may be configured to use the internal oscillator block as the system clock by performing one of the following actions:

- Program the FOSC<2:0> bits in Configuration Words to select the INTOSC clock source, which will be used as the default system clock upon a device Reset.
- Write the SCS<1:0> bits in the OSCCON register to switch the system clock source to the internal oscillator during run-time. See Section 5.3 "Clock Switching" for more information.

In **INTOSC** mode, OSC1/CLKIN is available for general purpose I/O. OSC2/CLKOUT is available for general purpose I/O or CLKOUT.

The function of the OSC2/CLKOUT pin is determined by the CLKOUTEN bit in Configuration Words.

The internal oscillator block has two independent oscillators and a dedicated Phase Lock Loop, HFPLL, that can produce one of three internal system clock sources.

- The HFINTOSC (High-Frequency Internal Oscillator) is factory calibrated and operates at 16 MHz. The HFINTOSC source is generated from the 500 kHz MFINTOSC source and the dedicated Phase Lock Loop, HFPLL. The frequency of the HFINTOSC can be user-adjusted via software using the OSCTUNE register (Register 5-3).
- 2. The **MFINTOSC** (Medium-Frequency Internal Oscillator) is factory calibrated and operates at 500 kHz. The frequency of the MFINTOSC can be user-adjusted via software using the OSCTUNE register (Register 5-3).
- 3. The **LFINTOSC** (Low-Frequency Internal Oscillator) is uncalibrated and operates at 31 kHz.

5.2.2.1 HFINTOSC

The High-Frequency Internal Oscillator (HFINTOSC) is a factory calibrated 16 MHz internal clock source. The frequency of the HFINTOSC can be altered via software using the OSCTUNE register (Register 5-3).

The output of the HFINTOSC connects to a postscaler and multiplexer (see Figure 5-1). One of multiple frequencies derived from the HFINTOSC can be selected via software using the IRCF<3:0> bits of the OSCCON register. See Section 5.2.2.7 "Internal Oscillator Clock Switch Timing" for more information.

The HFINTOSC is enabled by:

- Configure the IRCF<3:0> bits of the OSCCON register for the desired HF frequency, and
- FOSC<2:0> = 100, or
- Set the System Clock Source (SCS) bits of the OSCCON register to '1x'.

A fast start-up oscillator allows internal circuits to power up and stabilize before switching to HFINTOSC.

The High-Frequency Internal Oscillator Ready bit (HFIOFR) of the OSCSTAT register indicates when the HFINTOSC is running.

The High-Frequency Internal Oscillator Status Locked bit (HFIOFL) of the OSCSTAT register indicates when the HFINTOSC is running within 2% of its final value.

The High-Frequency Internal Oscillator Stable bit (HFIOFS) of the OSCSTAT register indicates when the HFINTOSC is running within 0.5% of its final value.

5.2.2.2 MFINTOSC

The Medium-Frequency Internal Oscillator (MFINTOSC) is a factory calibrated 500 kHz internal clock source. The frequency of the MFINTOSC can be altered via software using the OSCTUNE register (Register 5-3).

The output of the MFINTOSC connects to a postscaler and multiplexer (see Figure 5-1). One of nine frequencies derived from the MFINTOSC can be selected via software using the IRCF<3:0> bits of the OSCCON register. See Section 5.2.2.7 "Internal Oscillator Clock Switch Timing" for more information.

The MFINTOSC is enabled by:

- Configure the IRCF<3:0> bits of the OSCCON register for the desired HF frequency, and
- FOSC<2:0> = 100, or
- Set the System Clock Source (SCS) bits of the OSCCON register to '1x'.

The Medium-Frequency Internal Oscillator Ready bit (MFIOFR) of the OSCSTAT register indicates when the MFINTOSC is running.

5.2.2.3 Internal Oscillator Frequency Adjustment

The 500 kHz internal oscillator is factory calibrated. This internal oscillator can be adjusted in software by writing to the OSCTUNE register (Register 5-3). Since the HFINTOSC and MFINTOSC clock sources are derived from the 500 kHz internal oscillator, a change in the OSCTUNE register value will apply to both.

The default value of the OSCTUNE register is '0'. The value is a 6-bit two's complement number. A value of 1Fh will provide an adjustment to the maximum frequency. A value of 20h will provide an adjustment to the minimum frequency.

When the OSCTUNE register is modified, the oscillator frequency will begin shifting to the new frequency. Code execution continues during this shift. There is no indication that the shift has occurred.

OSCTUNE does not affect the LFINTOSC frequency. Operation of features that depend on the LFINTOSC clock source frequency, such as the Power-up Timer (PWRT), Watchdog Timer (WDT), Fail-Safe Clock Monitor (FSCM) and peripherals, are *not* affected by the change in frequency.

5.2.2.4 LFINTOSC

The Low-Frequency Internal Oscillator (LFINTOSC) is an uncalibrated 31 kHz internal clock source.

The output of the LFINTOSC connects to a multiplexer (see Figure 5-1). Select 31 kHz, via software, using the IRCF<3:0> bits of the OSCCON register. See Section 5.2.2.7 "Internal Oscillator Clock Switch Timing" for more information. The LFINTOSC is also the frequency for the Power-up Timer (PWRT), Watchdog Timer (WDT) and Fail-Safe Clock Monitor (FSCM).

The LFINTOSC is enabled by selecting 31 kHz (IRCF<3:0> bits of the OSCCON register = 000) as the system clock source (SCS bits of the OSCCON register = 1x), or when any of the following are enabled:

- Configure the IRCF<3:0> bits of the OSCCON register for the desired LF frequency, and
- FOSC<2:0> = 100, or
- Set the System Clock Source (SCS) bits of the OSCCON register to '1x'.

Peripherals that use the LFINTOSC are:

- Power-up Timer (PWRT)
- Watchdog Timer (WDT)
- Fail-Safe Clock Monitor (FSCM)

The Low-Frequency Internal Oscillator Ready bit (LFIOFR) of the OSCSTAT register indicates when the LFINTOSC is running.

5.2.2.5 Internal Oscillator Frequency Selection

The system clock speed can be selected via software using the Internal Oscillator Frequency Select bits IRCF<3:0> of the OSCCON register.

The postscaled output of the 16 MHz HFINTOSC, 500 kHz MFINTOSC, and 31 kHz LFINTOSC connect to a multiplexer (see Figure 5-1). The Internal Oscillator Frequency Select bits IRCF<3:0> of the OSCCON register select the frequency output of the internal oscillators. One of the following frequencies can be selected via software:

- 32 MHz (requires 4x PLL)
- 16 MHz
- 8 MHz
- 4 MHz
- 2 MHz
- 1 MHz
- 500 kHz (default after Reset)
- 250 kHz
- 125 kHz
- 62.5 kHz
- 31.25 kHz
- 31 kHz (LFINTOSC)
- Note: Following any Reset, the IRCF<3:0> bits of the OSCCON register are set to '0111' and the frequency selection is set to 500 kHz. The user can modify the IRCF bits to select a different frequency.

The IRCF<3:0> bits of the OSCCON register allow duplicate selections for some frequencies. These duplicate choices can offer system design trade-offs. Lower power consumption can be obtained when changing oscillator sources for a given frequency. Faster transition times can be obtained between frequency changes that use the same oscillator source.

5.2.2.6 32 MHz Internal Oscillator Frequency Selection

The Internal Oscillator Block can be used with the 4x PLL associated with the External Oscillator Block to produce a 32 MHz internal system clock source. The following settings are required to use the 32 MHz internal clock source:

- The FOSC bits in Configuration Words must be set to use the INTOSC source as the device system clock (FOSC<2:0> = 100).
- The SCS bits in the OSCCON register must be cleared to use the clock determined by FOSC<2:0> in Configuration Words (SCS<1:0> = 00).
- The IRCF bits in the OSCCON register must be set to the 8 MHz HFINTOSC set to use (IRCF<3:0> = 1110).
- The SPLLEN bit in the OSCCON register must be set to enable the 4x PLL, or the PLLEN bit of the Configuration Words must be programmed to a '1'.
 - **Note:** When using the PLLEN bit of the Configuration Words, the 4x PLL cannot be disabled by software and the SPLLEN option will not be available.

The 4x PLL is not available for use with the internal oscillator when the SCS bits of the OSCCON register are set to '1x'. The SCS bits must be set to '00' to use the 4x PLL with the internal oscillator.

5.2.2.7 Internal Oscillator Clock Switch Timing

When switching between the HFINTOSC, MFINTOSC and the LFINTOSC, the new oscillator may already be shut down to save power (see Figure 5-7). If this is the case, there is a delay after the IRCF<3:0> bits of the OSCCON register are modified before the frequency selection takes place. The OSCSTAT register will reflect the current active status of the HFINTOSC, MFINTOSC and LFINTOSC oscillators. The sequence of a frequency selection is as follows:

- 1. IRCF<3:0> bits of the OSCCON register are modified.
- 2. If the new clock is shut down, a clock start-up delay is started.
- 3. Clock switch circuitry waits for a falling edge of the current clock.
- 4. The current clock is held low and the clock switch circuitry waits for a rising edge in the new clock.
- 5. The new clock is now active.
- 6. The OSCSTAT register is updated as required.
- 7. Clock switch is complete.

See Figure 5-7 for more details.

If the internal oscillator speed is switched between two clocks of the same source, there is no start-up delay before the new frequency is selected. Clock switching time delays are shown in Table 5-1.

Start-up delay specifications are located in the oscillator tables of **Section 36.0** "Electrical **Specifications**".

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FIGURE 5-7:	INTERNAL OSCILLATOR SWITCH TIMING
HFINTOSC/→ LI	FINTOSC (FSCM and WDT disabled)
MFINTOSC	
HFINTOSC/ MFINTOSC	Start-up Time 2-cycle Sync Running
LFINTOSC	
IRCF <3:0>	$\neq 0$ $X = 0$
System Clock	
HFINTOSC/→ LF MFINTOSC	FINTOSC (Either FSCM or WDT enabled)
HFINTOSC/ MFINTOSC	2-cycle Sync Running
LFINTOSC -	
IRCF <3:0>	$\neq 0$ $X = 0$
System Clock	
LFINTOSC -+ HI	FINTOSC/MFINTOSC LFINTOSC turns off unless WDT or FSCM is enabled
LFINTOSC	
	Start-up Time 2-cycle Sync Running
HFINTOSC/ MFINTOSC	
IRCF <3:0>	= 0 × ≠ 0
System Clock	

5.3 Clock Switching

The system clock source can be switched between external and internal clock sources via software using the System Clock Select (SCS) bits of the OSCCON register. The following clock sources can be selected using the SCS bits:

- Default system oscillator determined by FOSC bits in Configuration Words
- Timer1 32 kHz crystal oscillator
- Internal Oscillator Block (INTOSC)

5.3.1 SYSTEM CLOCK SELECT (SCS) BITS

The System Clock Select (SCS) bits of the OSCCON register select the system clock source that is used for the CPU and peripherals.

- When the SCS bits of the OSCCON register = 00, the system clock source is determined by the value of the FOSC<2:0> bits in the Configuration Words.
- When the SCS bits of the OSCCON register = 01, the system clock source is the secondary oscillator.
- When the SCS bits of the OSCCON register = 1x, the system clock source is chosen by the internal oscillator frequency selected by the IRCF<3:0> bits of the OSCCON register. After a Reset, the SCS bits of the OSCCON register are always cleared.
 - Note: Any automatic clock switch, which may occur from Two-Speed Start-up or Fail-Safe Clock Monitor, does not update the SCS bits of the OSCCON register. The user can monitor the OSTS bit of the OSCSTAT register to determine the current system clock source.

When switching between clock sources, a delay is required to allow the new clock to stabilize. These oscillator delays are shown in Table 5-1.

5.3.2 OSCILLATOR START-UP TIMER STATUS (OSTS) BIT

The Oscillator Start-up Timer Status (OSTS) bit of the OSCSTAT register indicates whether the system clock is running from the external clock source, as defined by the FOSC<2:0> bits in the Configuration Words, or from the internal clock source. In particular, OSTS indicates that the Oscillator Start-up Timer (OST) has timed out for LP, XT or HS modes. The OST does not reflect the status of the secondary oscillator.

5.3.3 SECONDARY OSCILLATOR

The secondary oscillator is a separate crystal oscillator associated with the Timer1 peripheral. It is optimized for timekeeping operations with a 32.768 kHz crystal connected between the SOSCO and SOSCI device pins.

The secondary oscillator is enabled using the OSCEN control bit in the T1CON register. See Section 22.0 "Timer1/3/5 Module with Gate Control" for more information about the Timer1 peripheral.

5.3.4 SECONDARY OSCILLATOR READY (SOSCR) BIT

The user must ensure that the secondary oscillator is ready to be used before it is selected as a system clock source. The Secondary Oscillator Ready (SOSCR) bit of the OSCSTAT register indicates whether the secondary oscillator is ready to be used. After the SOSCR bit is set, the SCS bits can be configured to select the secondary oscillator.

5.3.5 CLOCK SWITCH BEFORE SLEEP

When a clock switch from an old clock to a new clock is requested just prior to entering Sleep mode, it is necessary to confirm that the switch is complete before the sleep instruction is executed. Failure to do so may result in an incomplete switch and consequential loss of the system clock altogether. Clock switching is confirmed by monitoring the clock status bits in the OSCSTAT register. Switch confirmation can be accomplished by sensing that the ready bit for the new clock is set or the ready bit for the old clock is cleared. For example, when switching between the internal oscillator with the PLL and the internal oscillator without the PLL, monitor the PLLR bit. When PLLR is set, the switch to 32 MHz operation is complete. Conversely, when PLLR is cleared the switch from 32 MHz operation to the selected internal clock is complete.

5.4 Two-Speed Clock Start-up Mode

Two-Speed Start-up mode provides additional power savings by minimizing the latency between external oscillator start-up and code execution. In applications that make heavy use of the Sleep mode, Two-Speed Start-up will remove the external oscillator start-up time from the time spent awake and can reduce the overall power consumption of the device. This mode allows the application to wake-up from Sleep, perform a few instructions using the INTOSC internal oscillator block as the clock source and go back to Sleep without waiting for the external oscillator to become stable.

Two-Speed Start-up provides benefits when the oscillator module is configured for LP, XT or HS modes. The Oscillator Start-up Timer (OST) is enabled for these modes and must count 1024 oscillations before the oscillator can be used as the system clock source.

If the oscillator module is configured for any mode other than LP, XT or HS mode, then Two-Speed Start-up is disabled. This is because the external clock oscillator does not require any stabilization time after POR or an exit from Sleep.

If the OST count reaches 1024 before the device enters Sleep mode, the OSTS bit of the OSCSTAT register is set and program execution switches to the external oscillator. However, the system may never operate from the external oscillator if the time spent awake is very short.

Note:	Executing a SLEEP instruction will abort
	the oscillator start-up time and will cause
	the OSTS bit of the OSCSTAT register to
	remain clear.

TABLE 5-1: OSCILLATOR SWITCHING DELAYS

Switch From	Switch To	Frequency	Oscillator Delay
Sleep	LFINTOSC ⁽¹⁾ MFINTOSC ⁽¹⁾ HFINTOSC ⁽¹⁾	31 kHz 31.25 kHz-500 kHz 31.25 kHz-16 MHz	Oscillator Warm-up Delay (Twarm) ⁽²⁾
Sleep/POR	EC, RC ⁽¹⁾	DC – 32 MHz	2 cycles
LFINTOSC	EC, RC ⁽¹⁾	DC – 32 MHz	1 cycle of each
Sleep/POR	Secondary Oscillator LP, XT, HS ⁽¹⁾	32 kHz-20 MHz	1024 Clock Cycles (OST)
Any clock source	MFINTOSC ⁽¹⁾ HFINTOSC ⁽¹⁾	31.25 kHz-500 kHz 31.25 kHz-16 MHz	2 μs (approx.)
Any clock source	LFINTOSC ⁽¹⁾	31 kHz	1 cycle of each
Any clock source	Secondary Oscillator	32 kHz	1024 Clock Cycles (OST)
PLL inactive	PLL active	16-32 MHz	2 ms (approx.)

Note 1: PLL inactive.

2: See Section 36.0 "Electrical Specifications".

5.4.1 TWO-SPEED START-UP MODE CONFIGURATION

Two-Speed Start-up mode is configured by the following settings:

- IESO (of the Configuration Words) = 1; Internal/External Switchover bit (Two-Speed Start-up mode enabled).
- SCS (of the OSCCON register) = 00.
- FOSC<2:0> bits in the Configuration Words configured for LP, XT or HS mode.

Two-Speed Start-up mode is entered after:

- Power-on Reset (POR) and, if enabled, after Power-up Timer (PWRT) has expired, or
- Wake-up from Sleep.

5.4.2 TWO-SPEED START-UP SEQUENCE

- 1. Wake-up from Power-on Reset or Sleep.
- 2. Instructions begin execution by the internal oscillator at the frequency set in the IRCF<3:0> bits of the OSCCON register.
- 3. OST enabled to count 1024 clock cycles.
- 4. OST timed out, wait for falling edge of the internal oscillator.
- 5. OSTS is set.
- 6. System clock held low until the next falling edge of new clock (LP, XT or HS mode).
- 7. System clock is switched to external clock source.

FIGURE 5-8: TWO-SPEED START-UP

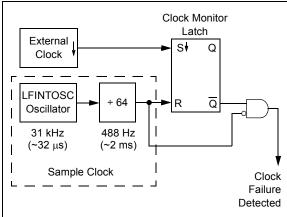
5.4.3 CHECKING TWO-SPEED CLOCK STATUS

Checking the state of the OSTS bit of the OSCSTAT register will confirm if the microcontroller is running from the external clock source, as defined by the FOSC<2:0> bits in the Configuration Words, or the internal oscillator.

5.5 Fail-Safe Clock Monitor

The Fail-Safe Clock Monitor (FSCM) allows the device to continue operating should the external oscillator fail. The FSCM can detect oscillator failure any time after the Oscillator Start-up Timer (OST) has expired. The FSCM is enabled by setting the FCMEN bit in the Configuration Words. The FSCM is applicable to all external Oscillator modes (LP, XT, HS, EC, Secondary Oscillator and RC).

FIGURE 5-9: FSCM BLOCK DIAGRAM



5.5.1 FAIL-SAFE DETECTION

The FSCM module detects a failed oscillator by comparing the external oscillator to the FSCM sample clock. The sample clock is generated by dividing the LFINTOSC by 64. See Figure 5-9. Inside the fail detector block is a latch. The external clock sets the latch on each falling edge of the external clock. The sample clock clears the latch on each rising edge of the sample clock. A failure is detected when an entire half-cycle of the sample clock elapses before the external clock goes low.

5.5.2 FAIL-SAFE OPERATION

When the external clock fails, the FSCM switches the device clock to an internal clock source and sets the bit flag OSFIF of the PIR2 register. Setting this flag will generate an interrupt if the OSFIE bit of the PIE2 register is also set. The device firmware can then take steps to mitigate the problems that may arise from a failed clock. The system clock will continue to be sourced from the internal clock source until the device firmware successfully restarts the external oscillator and switches back to external operation.

The internal clock source chosen by the FSCM is determined by the IRCF<3:0> bits of the OSCCON register. This allows the internal oscillator to be configured before a failure occurs.

5.5.3 FAIL-SAFE CONDITION CLEARING

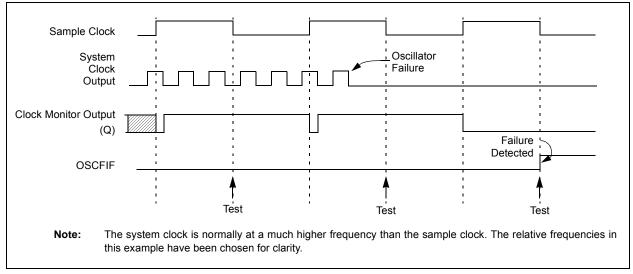
The Fail-Safe condition is cleared after a Reset, executing a SLEEP instruction or changing the SCS bits of the OSCCON register. When the SCS bits are changed, the OST is restarted. While the OST is running, the device continues to operate from the INTOSC selected in OSCCON. When the OST times out, the Fail-Safe condition is cleared after successfully switching to the external clock source. The OSFIF bit should be cleared prior to switching to the external clock source. If the Fail-Safe condition still exists, the OSFIF flag will again become set by hardware.

5.5.4 RESET OR WAKE-UP FROM SLEEP

The FSCM is designed to detect an oscillator failure after the Oscillator Start-up Timer (OST) has expired. The OST is used after waking up from Sleep and after any type of Reset. The OST is not used with the EC or RC Clock modes so that the FSCM will be active as soon as the Reset or wake-up has completed. When the FSCM is enabled, the Two-Speed Start-up is also enabled. Therefore, the device will always be executing code while the OST is operating.

Note: Due to the wide range of oscillator start-up times, the Fail-Safe circuit is not active during oscillator start-up (i.e., after exiting Reset or Sleep). After an appropriate amount of time, the user should check the Status bits in the OSCSTAT register to verify the oscillator start-up and that the system clock switchover has successfully completed.





5.6 Register Definitions: Oscillator Control

REGISTER 5-1: OSCCON: OSCILLATOR CONTROL REGISTER

R/W-0/0) R/W-0/0	R/W-1/1	R/W-1/1	R/W-1/1	U-0	R/W-0/0	R/W-0/0
SPLLEN	N	IRCF	<3:0>		_	SCS	<1:0>
bit 7							bit 0
Legend:							
R = Reada	able bit	W = Writable	bit	U = Unimplen	nented bit. rea	d as '0'	
u = Bit is u		x = Bit is unkr		•		DR/Value at all o	other Resets
'1' = Bit is s	•	'0' = Bit is cle					
bit 7	$\frac{\text{If PLLEN in }}{\text{SPLLEN bit}}$ $\frac{\text{If PLLEN in }}{1 = 4x \text{ PLL}}$	Configuration W	′ <u>ords = 1</u> : LL is always e	enabled (subject	to oscillator re	equirements)	
bit 6-3	$0 = 4x PLL \text{ is disabled}$ -3 IRCF<3:0>: Internal Oscillator Frequency Select bits $1111 = 16 \text{ MHz HF}$ $1110 = 8 \text{ MHz or 32 MHz HF}^{(2)}$ $1101 = 4 \text{ MHz HF}$ $1100 = 2 \text{ MHz HF}$ $1010 = 2 \text{ MHz HF}$ $1011 = 1 \text{ MHz HF}$ $1010 = 500 \text{ kHz HF}^{(1)}$ $1001 = 250 \text{ kHz HF}^{(1)}$ $1000 = 125 \text{ kHz HF}^{(1)}$ $0111 = 500 \text{ kHz MF} (default upon Reset)$ $0110 = 250 \text{ kHz MF}$ $0101 = 125 \text{ kHz MF}$ $0101 = 31.25 \text{ kHz HF}^{(1)}$ $0110 = 31.25 \text{ kHz MF}$						
bit 2	Unimpleme	nted: Read as '	0'				
bit 1-0	1x = Interna 01 = Second	System Clock S I oscillator block dary oscillator determined by F	(Configuration W	/ords		
	Duplicate frequer	ncy derived from	HFINTOSC.			Oscillator Fra	anonen.

2: 32 MHz when SPLLEN bit is set. Refer to Section 5.2.2.6 "32 MHz Internal Oscillator Frequency Selection".

R-1/q	R-0/q	R-q/q	R-0/q	R-0/q	R-q/q	R-0/0	R-0/q
SOSCR	PLLR	OSTS	HFIOFR	HFIOFL	MFIOFR	LFIOFR	HFIOFS
bit 7	•						bit 0
Legend:							
R = Readable	e bit	W = Writable	bit	U = Unimpler	nented bit, read	l as '0'	
u = Bit is unc		x = Bit is unk		•	at POR and BO		other Resets
'1' = Bit is set	•	'0' = Bit is cle		q = Condition	al		
bit 7	SOSCR: Se	condary Oscilla	tor Ready bit				
	If T1OSCEN						
		dary oscillator is					
		dary oscillator is	not ready				
	If T1OSCEN	<u>a – 0</u> . dary clock sourc	e is always rea	adv			
bit 6		L Ready bit		i di y			
bit o	1 = 4x PLL	•					
	0 = 4x PLL	is not ready					
bit 5	OSTS: Osci	illator Start-up T	imer Status bit				
		ng from the cloc				guration Word	S
		ng from an interr	•		00)		
bit 4		gh-Frequency I	nternal Oscillate	or Ready bit			
		DSC is ready					
		DSC is not read					
bit 3		gh-Frequency Ir		or Locked bit			
		DSC is at least 2 DSC is not 2% a					
bit 2		edium-Frequen		illator Ready b	it		
5112		OSC is ready			it i		
		OSC is not read	у				
bit 1	LFIOFR: Lo	w-Frequency In	ternal Oscillato	r Ready bit			
	1 = LFINTC	SC is ready		-			
	0 = LFINTC	OSC is not ready	/				
bit 0	HFIOFS: Hi	gh-Frequency li	nternal Oscillato	or Stable bit			
		OSC is at least (
	0 = HFINT(DSC is not 0.5%	accurate				

REGISTER 5-2: OSCSTAT: OSCILLATOR STATUS REGISTER

U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
_	_			TUN	<5:0>		
bit 7							bit 0
Legend:							
R = Readab	le bit	W = Writable	bit	U = Unimplen	nented bit, read	d as '0'	
u = Bit is un	changed	x = Bit is unkr	nown	-n/n = Value a	at POR and BC	R/Value at all	other Resets
'1' = Bit is se	et	'0' = Bit is clea	ared				
bit 7-6	Unimpleme	ented: Read as '	0'				
bit 5-0	TUN<5:0>:	Frequency Tunir	ng bits				
	100000 =	Minimum frequer	псу				
	•						
	•						
	• 1111111 =						
		Oscillator module	e is running at	the factory-calil	brated frequen	cv	
	000001 =		J	, ,		- 5	
	•						
	•						
	•						
	011110 =						

REGISTER 5-3: OSCTUNE: OSCILLATOR TUNING REGISTER

TABLE 5-2: SUMMARY OF REGISTERS ASSOCIATED WITH CLOCK SOURCES

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
OSCCON	SPLLEN		IRCF<3:0>				SCS	<1:0>	116
OSCSTAT	SOSCR	PLLR	OSTS	HFIOFR	HFIOFL	MFIOFR	LFIOFR	HFIOFS	117
OSCTUNE	_	_			TUN	<5:0>			118
PIR2	OSFIF	C2IF	C1IF	COG1IF	BCL1IF	C4IF	C3IF	CCP2IF	140
PIE2	OSFIE	C2IE	C1IE	COG1IE	BCL1IE	C4IE	C3IE	CCP2IE	134
T1CON	CS<	1:0>	CKPS	<1:0>	OSCEN	SYNC	_	ON	283

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by clock sources.

TABLE 5-3: SUMMARY OF CONFIGURATION WORD WITH CLOCK SOURCES

Name	Bits	Bit -/7	Bit -/6	Bit 13/5	Bit 12/4	Bit 11/3	Bit 10/2	Bit 9/1	Bit 8/0	Register on Page
	13:8	_	_	FCMEN	IESO	CLKOUTEN	BOREN<1:0>		_	05
CONFIG1	7:0	CP	MCLRE	PWRTE	WDT	ΓE<1:0>		FOSC<2:0>		95

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by clock sources.

A simplified block diagram of the On-Chip Reset Circuit

is shown in Figure 6-1.

6.0 RESETS

There are multiple ways to reset this device:

- Power-on Reset (POR)
- Brown-out Reset (BOR)
- Low-Power Brown-out Reset (LPBOR)
- MCLR Reset
- WDT Reset
- RESET instruction
- Stack Overflow
- Stack Underflow
- · Programming mode exit

To allow VDD to stabilize, an optional Power-up Timer can be enabled to extend the Reset time after a BOR or POR event.

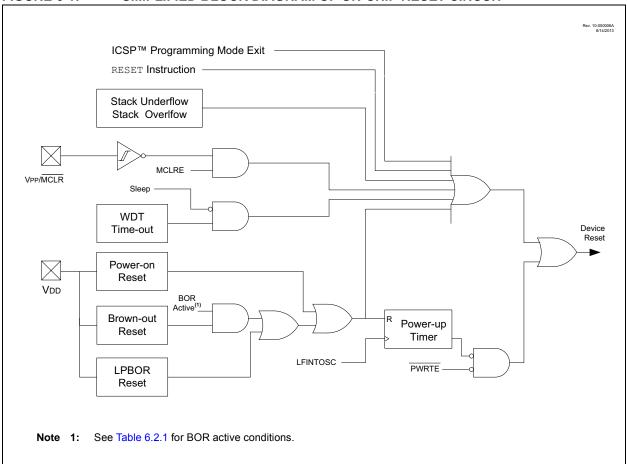


FIGURE 6-1: SIMPLIFIED BLOCK DIAGRAM OF ON-CHIP RESET CIRCUIT

6.1 Power-On Reset (POR)

The POR circuit holds the device in Reset until VDD has reached an acceptable level for minimum operation. Slow rising VDD, fast operating speeds or analog performance may require greater than minimum VDD. The PWRT, BOR or MCLR features can be used to extend the start-up period until all device operation conditions have been met.

6.1.1 POWER-UP TIMER (PWRT)

The Power-up Timer provides a nominal 64 ms time-out on POR or Brown-out Reset.

The device is held in Reset as long as PWRT is active. The PWRT delay allows additional time for the VDD to rise to an acceptable level. The Power-up Timer is enabled by clearing the PWRTE bit in Configuration Words.

The Power-up Timer starts after the release of the POR and BOR.

For additional information, refer to Application Note AN607, *"Power-up Trouble Shooting"* (DS00607).

6.2 Brown-Out Reset (BOR)

The BOR circuit holds the device in Reset when VDD reaches a selectable minimum level. Between the POR and BOR, complete voltage range coverage for execution protection can be implemented.

The Brown-out Reset module has four operating modes controlled by the BOREN<1:0> bits in Configuration Words. The four operating modes are:

- · BOR is always on
- · BOR is off when in Sleep
- · BOR is controlled by software
- · BOR is always off

TABLE 0-1:	DUR UPERATI	NG WIODES		
BOREN<1:0>	SBOREN	Device Mode	BOR Mode	Instruction Execution upon: Release of POR or Wake-up from Sleep
11	х	Х	Active	Waits for BOR ready ⁽¹⁾ (BORRDY = 1)
1.0	37	Awake	Active	White for POP ready (POPPDY = 1)
10	Х	Sleep	Disabled	Waits for BOR ready (BORRDY = 1)
0.1	1	х	Active	Waits for BOR ready ⁽¹⁾ (BORRDY = 1)
01	0	х	Disabled	Paging immediately (PORPDY =)
00	х	х	Disabled	Begins immediately (BORRDY = x)

TABLE 6-1:BOR OPERATING MODES

Refer to Table 6.2.1 for more information.

The Brown-out Reset voltage level is selectable by configuring the BORV bit in Configuration Words.

A VDD noise rejection filter prevents the BOR from triggering on small events. If VDD falls below VBOR for a duration greater than parameter TBORDC, the device will reset. See Figure 6-2 for more information.

6.2.1 BOR IS ALWAYS ON

When the BOREN bits of Configuration Words are programmed to '11', the BOR is always on. The device start-up will be delayed until the BOR is ready and VDD is higher than the BOR threshold.

BOR protection is active during Sleep. The BOR does not delay wake-up from Sleep.

6.2.2 BOR IS OFF IN SLEEP

When the BOREN bits of Configuration Words are programmed to '10', the BOR is on, except in Sleep. The device start-up will be delayed until the BOR is ready and VDD is higher than the BOR threshold.

BOR protection is not active during Sleep. The device wake-up will be delayed until the BOR is ready.

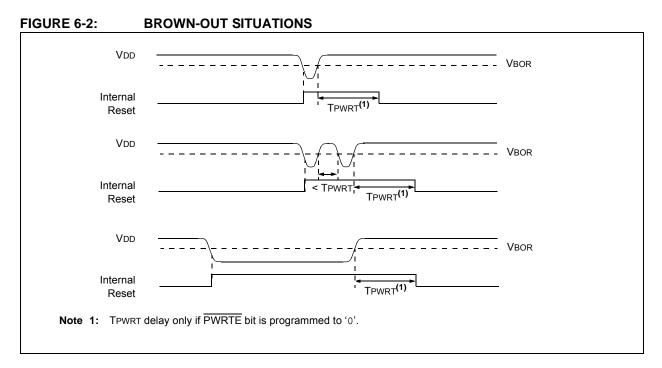
6.2.3 BOR CONTROLLED BY SOFTWARE

When the BOREN bits of Configuration Words are programmed to '01', the BOR is controlled by the SBOREN bit of the BORCON register. The device start-up is not delayed by the BOR ready condition or the VDD level.

BOR protection begins as soon as the BOR circuit is ready. The status of the BOR circuit is reflected in the BORRDY bit of the BORCON register.

BOR protection is unchanged by Sleep.

Note 1: In these specific cases, "Release of POR" and "Wake-up from Sleep", there is no delay in start-up. The BOR ready flag, (BORRDY = 1), will be set before the CPU is ready to execute instructions because the BOR circuit is forced on by the BOREN<1:0> bits.



6.3 Register Definitions: BOR Control

REGISTER 6-1: BORCON: BROWN-OUT RESET CONTROL REGISTER

R/W-1/u	R/W-0/u	U-0	U-0	U-0	U-0	U-0	R-q/u
SBOREN	BORFS ⁽¹⁾	—	—	—	—	—	BORRDY
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 7	SBOREN: Software Brown-out Reset Enable bit If BOREN <1:0> in Configuration Words ≠ 01: SBOREN is read/write, but has no effect on the BOR. If BOREN <1:0> in Configuration Words = 01: 1 = BOR Enabled 0 = BOR Disabled
bit 6	 BORFS: Brown-out Reset Fast Start bit⁽¹⁾ If BOREN<1:0> = 11 (Always on) or BOREN<1:0> = 00 (Always off) BORFS is Read/Write, but has no effect. If BOREN<1:0> = 10 (Disabled in Sleep) or BOREN<1:0> = 01 (Under software control): 1 = Band gap is forced on always (covers sleep/wake-up/operating cases) 0 = Band gap operates normally, and may turn off
bit 5-1	Unimplemented: Read as '0'
bit 0	BORRDY: Brown-out Reset Circuit Ready Status bit 1 = The Brown-out Reset circuit is active 0 = The Brown-out Reset circuit is inactive

Note 1: BOREN<1:0> bits are located in Configuration Words.

6.4 Low-Power Brown-Out Reset (LPBOR)

The Low-Power Brown-out Reset (LPBOR) is an essential part of the Reset subsystem. Refer to Figure 6-1 to see how the BOR interacts with other modules.

The LPBOR is used to monitor the external VDD pin. When too low of a voltage is detected, the device is held in Reset. When this occurs, a register bit ($\overline{\text{BOR}}$) is changed to indicate that a BOR Reset has occurred. The same bit is set for both the BOR and the LPBOR. Refer to Register 6-2.

6.4.1 ENABLING LPBOR

The LPBOR is controlled by the LPBOR bit of Configuration Words. When the device is erased, the LPBOR module defaults to disabled.

6.4.1.1 LPBOR Module Output

The output of the LPBOR module is a signal indicating whether or not a Reset is to be asserted. This signal is OR'd together with the Reset signal of the BOR module to provide the generic BOR signal, which goes to the PCON register and to the power control block.

6.5 MCLR

The $\overline{\text{MCLR}}$ is an <u>optional</u> external input that can reset the device. The $\overline{\text{MCLR}}$ function is controlled by the MCLRE bit of Configuration Words and the LVP bit of Configuration Words (Table 6-2).

TABLE 6-2: MCLR CONFIGURATION

MCLRE	LVP	MCLR
x	1	Enabled
1	0	Enabled
0	0	Disabled

6.5.1 MCLR ENABLED

When MCLR is enabled and the pin is held low, the device is held in Reset. The MCLR pin is connected to VDD through an internal weak pull-up.

The device has a noise filter in the $\overline{\text{MCLR}}$ Reset path. The filter will detect and ignore small pulses.

Note: A Reset does not drive the MCLR pin low.

6.5.2 MCLR DISABLED

When MCLR is disabled, the pin functions as a general purpose input and the internal weak pull-up is under software control. See **Section 11.1** "**PORTA Registers**" for more information.

6.6 Watchdog Timer (WDT) Reset

The Watchdog Timer generates a Reset if the firmware does not issue a CLRWDT instruction within the time-out period. The TO and PD bits in the STATUS register are changed to indicate the WDT Reset. See **Section 9.0** "**Watchdog Timer (WDT)**" for more information.

6.7 RESET Instruction

A RESET instruction will cause a device Reset. The \overline{RI} bit in the PCON register will be set to '0'. See Table 6-4 for default conditions after a RESET instruction has occurred.

6.8 Stack Overflow/Underflow Reset

The device can reset when the Stack Overflows or Underflows. The STKOVF or STKUNF bits of the PCON register indicate the Reset condition. These Resets are enabled by setting the STVREN bit in Configuration Words. See **Section 3.6.2** "Overflow/Underflow **Reset**" for more information.

6.9 Programming Mode Exit

Upon exit of Programming mode, the device will behave as if a POR had just occurred.

6.10 Power-Up Timer

The Power-up Timer optionally delays device execution after a BOR or POR event. This timer is typically used to allow VDD to stabilize before allowing the device to start running.

The Power-up Timer is controlled by the PWRTE bit of Configuration Words.

6.11 Start-up Sequence

Upon the release of a POR or BOR, the following must occur before the device will begin executing:

- 1. Power-up Timer runs to completion (if enabled).
- 2. Oscillator start-up timer runs to completion (if required for oscillator source).
- 3. MCLR must be released (if enabled).

The total time-out will vary based on oscillator configuration and Power-up Timer configuration. See Section 5.0 "Oscillator Module (with Fail-Safe Clock Monitor)" for more information.

The Power-up Timer and oscillator start-up timer run independently of MCLR Reset. If MCLR is kept low long enough, the Power-up Timer and oscillator start-up timer will expire. Upon bringing MCLR high, the device will begin execution after 10 Fosc cycles (see Figure 6-3). This is useful for testing purposes or to synchronize more than one device operating in parallel.

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FIGURE 6-3:	RESET START-UP SEQUENCE
Internal POR	
Power-up Timer	
MCLR	
Internal RESET	
	Oscillator Modes
External Crystal	
Oscillator Start-up Timer	
Oscillator Fosc	
Internal Oscillator	
Oscillator	
Fosc	
External Clock (EC)	
CLKIN	
Fosc	

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6.12 Determining the Cause of a Reset

Upon any Reset, multiple bits in the STATUS and PCON registers are updated to indicate the cause of the Reset. Table 6-3 and Table 6-4 show the Reset conditions of these registers.

STKOVF	STKUNF	RWDT	RMCLR	RI	POR	BOR	то	PD	Condition
0	0	1	1	1	0	x	1	1	Power-on Reset
0	0	1	1	1	0	x	0	х	Illegal, $\overline{\text{TO}}$ is set on $\overline{\text{POR}}$
0	0	1	1	1	0	x	x	0	Illegal, \overline{PD} is set on \overline{POR}
0	0	u	1	1	u	0	1	1	Brown-out Reset
u	u	0	u	u	u	u	0	u	WDT Reset
u	u	u	u	u	u	u	0	0	WDT Wake-up from Sleep
u	u	u	u	u	u	u	1	0	Interrupt Wake-up from Sleep
u	u	u	0	u	u	u	u	u	MCLR Reset during normal operation
u	u	u	0	u	u	u	1	0	MCLR Reset during Sleep
u	u	u	u	0	u	u	u	u	RESET Instruction Executed
1	u	u	u	u	u	u	u	u	Stack Overflow Reset (STVREN = 1)
u	1	u	u	u	u	u	u	u	Stack Underflow Reset (STVREN = 1)

TABLE 6-3: RESET STATUS BITS AND THEIR SIGNIFICANCE

TABLE 6-4: RESET CONDITION FOR SPECIAL REGISTERS

Condition	Program Counter	STATUS Register	PCON Register
Power-on Reset	0000h	1 1000	00 110x
MCLR Reset during normal operation	0000h	u uuuu	uu Ouuu
MCLR Reset during Sleep	0000h	1 Ouuu	uu Ouuu
WDT Reset	0000h	0 uuuu	uu uuuu
WDT Wake-up from Sleep	PC + 1	0 Ouuu	uu uuuu
Brown-out Reset	0000h	1 luuu	00 11u0
Interrupt Wake-up from Sleep	PC + 1 ⁽¹⁾	1 Ouuu	uu uuuu
RESET Instruction Executed	0000h	u uuuu	uu u0uu
Stack Overflow Reset (STVREN = 1)	0000h	u uuuu	lu uuuu
Stack Underflow Reset (STVREN = 1)	0000h	u uuuu	ul uuuu

Legend: u = unchanged, x = unknown, - = unimplemented bit, reads as '0'.

Note 1: When the wake-up is due to an interrupt and Global Enable bit (GIE) is set, the return address is pushed on the stack and PC is loaded with the interrupt vector (0004h) after execution of PC + 1.

6.13 Power Control (PCON) Register

The Power Control (PCON) register contains flag bits to differentiate between a:

- Power-on Reset (POR)
- Brown-out Reset (BOR)
- Reset Instruction Reset (RI)
- MCLR Reset (RMCLR)
- Watchdog Timer Reset (RWDT)
- Stack Underflow Reset (STKUNF)
- Stack Overflow Reset (STKOVF)

6.14 Register Definitions: Power Control

REGISTER 6-2: PCON: POWER CONTROL REGISTER

R/W/HS-0/q	R/W/HS-0/q	U-0	R/W/HC-1/q	R/W/HC-1/q	R/W/HC-1/q	R/W/HC-q/u	R/W/HC-q/u
STKOVF	STKUNF	—	RWDT	RMCLR	RI	POR	BOR
bit 7	•						bit 0

Legend:		
HC = Bit is cleared by har	dware	HS = Bit is set by hardware
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-m/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 7	STKOVF: Stack Overflow Flag bit
	1 = A Stack Overflow occurred
	0 = A Stack Overflow has not occurred or cleared by firmware
bit 6	STKUNF: Stack Underflow Flag bit
	1 = A Stack Underflow occurred
	0 = A Stack Underflow has not occurred or cleared by firmware
bit 5	Unimplemented: Read as '0'
bit 4	RWDT: Watchdog Timer Reset Flag bit
	1 = A Watchdog Timer Reset has not occurred or set to '1' by firmware
	0 = A Watchdog Timer Reset has occurred (cleared by hardware)
bit 3	RMCLR: MCLR Reset Flag bit
	1 = A MCLR Reset has not occurred or set to '1' by firmware
	0 = A MCLR Reset has occurred (cleared by hardware)
bit 2	RI: RESET Instruction Flag bit
	1 = A RESET instruction has not been executed or set to '1' by firmware
	0 = A RESET instruction has been executed (cleared by hardware)
bit 1	POR: Power-on Reset Status bit
	1 = No Power-on Reset occurred
	0 = A Power-on Reset occurred (must be set in software after a Power-on Reset occurs)
bit 0	BOR: Brown-out Reset Status bit
	1 = No Brown-out Reset occurred
	0 = A Brown-out Reset occurred (must be set in software after a Power-on Reset or Brown-out Reset
	occurs)

The PCON register bits are shown in Register 6-2.

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Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page		
BORCON	SBOREN	BORFS	_		_			BORRDY	121		
PCON	STKOVF	STKUNF	_	RWDT	RMCLR	RI	POR	BOR	125		
STATUS			_	TO	PD	Z	DC	С	40		
WDTCON		_		V	VDTPS<4:0	>		SWDTEN	154		

TABLE 6-5: SUMMARY OF REGISTERS ASSOCIATED WITH RESETS

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by Resets.

7.0 INTERRUPTS

The interrupt feature allows certain events to preempt normal program flow. Firmware is used to determine the source of the interrupt and act accordingly. Some interrupts can be configured to wake the MCU from Sleep mode.

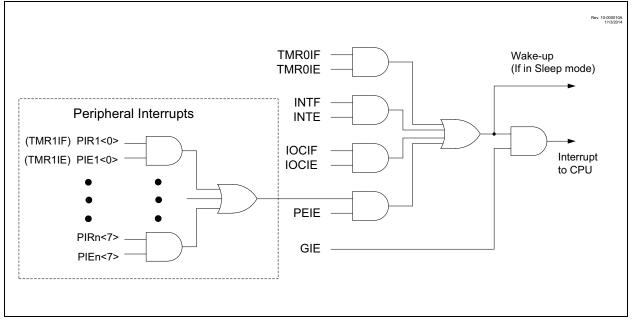
This chapter contains the following information for Interrupts:

- Operation
- Interrupt Latency
- Interrupts During Sleep
- INT Pin
- · Automatic Context Saving

Many peripherals produce interrupts. Refer to the corresponding chapters for details.

A block diagram of the interrupt logic is shown in Figure 7-1.





7.1 Operation

Interrupts are disabled upon any device Reset. They are enabled by setting the following bits:

- · GIE bit of the INTCON register
- Interrupt Enable bit(s) for the specific interrupt event(s)
- PEIE bit of the INTCON register (if the Interrupt Enable bit of the interrupt event is contained in the PIE1 or PIE2 registers)

The INTCON, PIR1 and PIR2 registers record individual interrupts via interrupt flag bits. Interrupt flag bits will be set, regardless of the status of the GIE, PEIE and individual interrupt enable bits.

The following events happen when an interrupt event occurs while the GIE bit is set:

- · Current prefetched instruction is flushed
- · GIE bit is cleared
- Current Program Counter (PC) is pushed onto the stack
- Critical registers are automatically saved to the shadow registers (See "Section 7.5 "Automatic Context Saving")
- · PC is loaded with the interrupt vector 0004h

The firmware within the Interrupt Service Routine (ISR) should determine the source of the interrupt by polling the interrupt flag bits. The interrupt flag bits must be cleared before exiting the ISR to avoid repeated interrupts. Because the GIE bit is cleared, any interrupt that occurs while executing the ISR will be recorded through its interrupt flag, but will not cause the processor to redirect to the interrupt vector.

The RETFIE instruction exits the ISR by popping the previous address from the stack, restoring the saved context from the shadow registers and setting the GIE bit.

For additional information on a specific interrupt's operation, refer to its peripheral chapter.

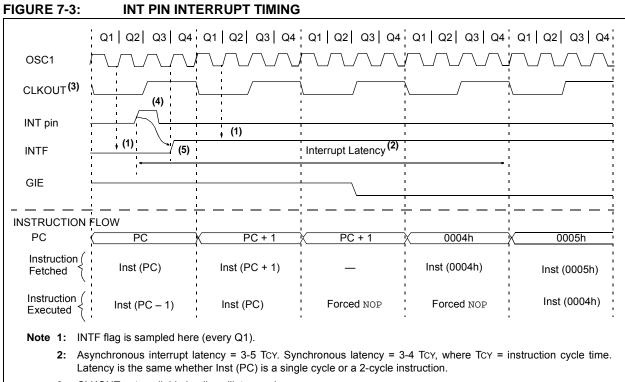
- Note 1: Individual interrupt flag bits are set, regardless of the state of any other enable bits.
 - 2: All interrupts will be ignored while the GIE bit is cleared. Any interrupt occurring while the GIE bit is clear will be serviced when the GIE bit is set again.

7.2 Interrupt Latency

Interrupt latency is defined as the time from when the interrupt event occurs to the time code execution at the interrupt vector begins. The latency for synchronous interrupts is three or four instruction cycles. For asynchronous interrupts, the latency is three to five instruction cycles, depending on when the interrupt occurs. See Figure 7-2 and Figure 7-3 for more details.

FIGURE 7	7-2:	INTERRUPT	LATENCY	,				
OSC1								
		4 Q1 Q2 Q3 Q4						Q1 Q2 Q3 Q4
CLKR			Interru during	upt Sampled Q1				
Interrupt								
GIE								
PC	PC-1	PC	PC	2+1	0004h	0005h		
Execute	1 Cycle Ins	truction at PC	Inst(PC)	NOP	NOP	Inst(0004h)		
Interrupt								
GIE								
PC	PC-1	PC	PC+1/FSR ADDR	New PC/ PC+1	0004h	0005h		
Execute-	2 Cycle Ins	struction at PC	Inst(PC)	NOP	NOP	Inst(0004h)		
Interrupt								
GIE								
PC	PC-1	PC	FSR ADDR	PC+1	PC+2	0004h	0005h	
Execute	3 Cycle Ins	truction at PC	INST(PC)	NOP	NOP	NOP	Inst(0004h)	Inst(0005h)
Interrupt								
GIE								
PC	PC-1	РС	FSR ADDR	PC+1	P	2+2	0004h	0005h
Execute	3 Cycle Ins	truction at PC	INST(PC)	NOP	NOP	NOP	NOP	Inst(0004h)

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3: CLKOUT not available in all oscillator modes.

4: For minimum width of INT pulse, refer to AC specifications in Section 36.0 "Electrical Specifications"".

5: INTF is enabled to be set any time during the Q4-Q1 cycles.

7.3 Interrupts During Sleep

Some interrupts can be used to wake from Sleep. To wake from Sleep, the peripheral must be able to operate without the system clock. The interrupt source must have the appropriate Interrupt Enable bit(s) set prior to entering Sleep.

On waking from Sleep, if the GIE bit is also set, the processor will branch to the interrupt vector. Otherwise, the processor will continue executing instructions after the SLEEP instruction. The instruction directly after the SLEEP instruction will always be executed before branching to the ISR. Refer to **Section 8.0** "**Power-Down Mode (Sleep)**" for more details.

7.4 INT Pin

The INT pin can be used to generate an asynchronous edge-triggered interrupt. This interrupt is enabled by setting the INTE bit of the INTCON register. The INTEDG bit of the OPTION_REG register determines on which edge the interrupt will occur. When the INTEDG bit is set, the rising edge will cause the interrupt. When the INTEDG bit is clear, the falling edge will cause the interrupt. The INTF bit of the INTCON register will be set when a valid edge appears on the INT pin. If the GIE and INTE bits are also set, the processor will redirect program execution to the interrupt vector.

7.5 Automatic Context Saving

Upon entering an interrupt, the return PC address is saved on the stack. Additionally, the following registers are automatically saved in the shadow registers:

- W register
- STATUS register (except for TO and PD)
- BSR register
- FSR registers
- PCLATH register

Upon exiting the Interrupt Service Routine, these registers are automatically restored. Any modifications to these registers during the ISR will be lost. If modifications to any of these registers are desired, the corresponding shadow register should be modified and the value will be restored when exiting the ISR. The shadow registers are available in Bank 31 and are readable and writable. Depending on the user's application, other registers may also need to be saved.

7.6 Register Definitions: Interrupt Control

REGISTER 7-1: INTCON: INTERRUPT CONTROL REGISTER

1 = Enables a 0 = Disables a PEIE: Periphe	TMR0IE W = Writable I x = Bit is unkn '0' = Bit is clea nterrupt Enable Il active interru all interrupts	own ared bit		TMR0IF nented bit, read at POR and BOI		IOCIF ⁽¹⁾ bit 0
GIE: Global Ir 1 = Enables a 0 = Disables a PEIE: Periphe	x = Bit is unkn '0' = Bit is clea nterrupt Enable Il active interru	own ared bit				
GIE: Global Ir 1 = Enables a 0 = Disables a PEIE: Periphe	x = Bit is unkn '0' = Bit is clea nterrupt Enable Il active interru	own ared bit				ther Resets
GIE: Global Ir 1 = Enables a 0 = Disables a PEIE: Periphe	x = Bit is unkn '0' = Bit is clea nterrupt Enable Il active interru	own ared bit				ther Resets
GIE: Global Ir 1 = Enables a 0 = Disables a PEIE: Periphe	x = Bit is unkn '0' = Bit is clea nterrupt Enable Il active interru	own ared bit				ther Resets
GIE: Global Ir 1 = Enables a 0 = Disables a PEIE: Periphe	'0' = Bit is clea nterrupt Enable Il active interru	bit	-n/n = Value a	at POR and BOI	R/Value at all o	ther Resets
1 = Enables a 0 = Disables a PEIE: Periphe	nterrupt Enable Il active interru	bit				
1 = Enables a 0 = Disables a PEIE: Periphe	Il active interru					
0 = Disables a PEIE: Periphe		pts				
	eral Interrupt Er Il active periph all peripheral in	eral interrupts	3			
1 = Enables th	er0 Overflow In ne Timer0 inter he Timer0 inter	rupt	e bit			
1 = Enables th	ne INT external	l interrupt				
1 = Enables th	ne interrupt-on-	-change				
1 = TMR0 reg	ister has overfl	lowed	bit			
1 = The INT e	xternal interrup	ot occurred	Jr			
1 = When at le	east one of the	interrupt-on-	change pins ch			
	NTE: INT Ext = Enables th = Disables t OCIE: Interru = Enables t = Disables t TMR0IF: Time = TMR0 reg = TMR0 reg = TMR0 reg NTF: INT Ext = The INT e = The INT e	NTE: INT External Interrupt = Enables the INT external = Disables the INT external OCIE: Interrupt-on-Change = Enables the interrupt-on = Disables the interrupt-on TMR0IF: Timer0 Overflow In = TMR0 register has overfl = TMR0 register has overfl = TMR0 register did not ov NTF: INT External Interrupt = The INT external interrupt	 = TMR0 register has overflowed = TMR0 register did not overflow NTF: INT External Interrupt Flag bit = The INT external interrupt occurred = The INT external interrupt did not occur OCIF: Interrupt-on-Change Interrupt Flag = When at least one of the interrupt-on- = None of the interrupt-on-change pins 	 NTE: INT External Interrupt Enable bit = Enables the INT external interrupt = Disables the INT external interrupt OCIE: Interrupt-on-Change Enable bit = Enables the interrupt-on-change = Disables the interrupt-on-change MROIF: Timer0 Overflow Interrupt Flag bit = TMR0 register has overflowed = TMR0 register did not overflow NTF: INT External Interrupt Flag bit = The INT external interrupt did not occur OCIF: Interrupt-on-Change Interrupt Flag bit⁽¹⁾ = When at least one of the interrupt-on-change pins have changed 	 NTE: INT External Interrupt Enable bit = Enables the INT external interrupt = Disables the INT external interrupt OCIE: Interrupt-on-Change Enable bit = Enables the interrupt-on-change = Disables the interrupt-on-change MROIF: Timer0 Overflow Interrupt Flag bit = TMR0 register has overflowed = TMR0 register did not overflow NTF: INT External Interrupt Flag bit = The INT external interrupt did not occur OCIF: Interrupt-on-Change Interrupt Flag bit⁽¹⁾ = When at least one of the interrupt-on-change pins changed state = None of the interrupt-on-change pins have changed state 	NTE: INT External Interrupt Enable bit = Enables the INT external interrupt OCIE: Interrupt-on-Change Enable bit = Enables the interrupt-on-change = Disables the interrupt-on-change TMROIF: Timer0 Overflow Interrupt Flag bit = TMR0 register has overflowed = TMR0 register did not overflow NTF: INT External Interrupt Flag bit = The INT external interrupt occurred = The INT external interrupt did not occur OCIF: Interrupt-on-Change Interrupt Flag bit ⁽¹⁾ = When at least one of the interrupt-on-change pins changed state

Note 1: The IOCIF Flag bit is read-only and cleared when all the interrupt-on-change flags in the IOCxF registers have been cleared by software.

Note: Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the Global Enable bit, GIE, of the INTCON register. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	CCP1IE	TMR2IE	TMR1IE
bit 7							bit C
Legend:							
R = Readable		W = Writable		•	mented bit, read		
u = Bit is unch	•	x = Bit is unkr		-n/n = Value a	at POR and BO	R/Value at all o	ther Resets
'1' = Bit is set		'0' = Bit is clea	ared				
bit 7	TMR1GIF: Ti	mer1 Gate Inte	rrupt Enable h	nit			
		he Timer1 gate	•				
		the Timer1 gate	•	•			
bit 6	ADIE: Analog	j-to-Digital Con	verter (ADC) I	nterrupt Enabl	e bit		
		he ADC interru					
		the ADC interru	•				
bit 5		RT Receive Inte	•				
		he EUSART re the EUSART re					
bit 4		RT Transmit Inte					
		he EUSART tra	-				
		the EUSART tra					
bit 3	SSP1IE: Syne	chronous Seria	I Port (MSSP)	Interrupt Enal	ble bit		
		he MSSP inter					
		the MSSP inter	•				
bit 2		P1 Interrupt Ena he CCP1 interr					
		the CCP1 inter					
bit 1		R2 to T2PR Ma	•	Enable bit			
		he Timer2 to T	•				
	0 = Disables	the Timer2 to T	2PR match in	terrupt			
bit 0	TMR1IE: Tim	er1 Overflow In	terrupt Enable	e bit			
		he Timer1 over					
	0 = Disables	the Timer1 ove	flow interrupt				
N. (TOON					
	PEIE of the IN	•					
se	t to enable any p		սրւ.				

REGISTER 7-2: PIE1: PERIPHERAL INTERRUPT ENABLE REGISTER 1

R/W-0/0	R/W-0/0	R/W-0/0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0				
OSFIE	C2IE	C1IE	COG1IE	BCL1IE	C4IE	C3IE	CCP2IE				
bit 7							bit				
Logondi											
Legend: R = Readable	e bit	W = Writable	bit	U = Unimpler	nented bit read	l as '0'					
	R = Readable bitW = Writable bitU = Unimplemented bit, read as '0'u = Bit is unknown-n/n = Value at POR and BOR/Value at all ot										
'1' = Bit is set	0	'0' = Bit is cle									
bit 7	OSFIE: Osci	llator Fail Interr	upt Enable bit								
		the Oscillator F the Oscillator									
bit 6	C2IE: Comp	arator C2 Interr	upt Enable bit								
		the Comparate									
		the Comparate	•								
bit 5	•	arator C1 Interr	•								
		the Comparato the Comparato									
bit 4		DG1 Auto-Shuto	•								
	1 = COG1 ir	nterrupt enableo nterrupt disable	t								
bit 3		BCL1IE: MSSP Bus Collision Interrupt Enable bit									
	1 = Enables the MSSP Bus Collision interrupt										
	0 = Disables	s the MSSP Bus	s Collision inte	errupt							
bit 2	C4IE: TMR6	to T6PR Match	Interrupt Ena	ıble bit							
		 Enables the Comparator C4 interrupt Disables the Comparator C4 interrupt 									
bit 1		•	•								
DILI		: TMR4 to T4PR Match Interrupt Enable bit Enables the Comparator C3 interrupt									
		the Comparate									
bit 0	CCP2IE: CC	P2 Interrupt En	able bit								
	1 = Enables	the CCP2 inter	rupt								
	0 = Disables	s the CCP2 inte	rrupt								
Note: Bit		ITCON register	must be								
		peripheral inter									

REGISTER 7-3: PIE2: PERIPHERAL INTERRUPT ENABLE REGISTER 2

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0				
_	_	COG2IE	ZCDIE	CLC4IE	CLC3IE	CLC2IE	CLC1IE				
bit 7							bit 0				
Legend:											
R = Reada	ble bit	W = Writable	bit	•	nented bit, read						
u = Bit is u	nchanged	x = Bit is unkr	nown	-n/n = Value a	at POR and BO	R/Value at all o	ther Resets				
'1' = Bit is s	set	'0' = Bit is cle	ared								
1.11 7.0			o.1								
bit 7-6	-	nted: Read as '									
bit 5		DG2 Auto-Shutd	•	Enable bit							
		nterrupt enabled nterrupt disabled									
bit 4		E: Zero-Cross Detection Interrupt Enable bit									
	1 = ZCD inte	1 = ZCD interrupt enabled									
	0 = ZCD integrated on the second s	0 = ZCD interrupt disabled									
bit 3		: CLC4 Interrupt Enable bit									
		terrupt enabled	1								
bit 2		0 = CLC4 interrupt disabled									
		CLC3IE: CLC3 Interrupt Enable bit 1 = CLC3 interrupt enabled									
		terrupt disabled									
bit 1	CLC2IE: CL	C2 Interrupt Ena	able bit								
		1 = CLC2 interrupt enabled									
		terrupt disabled									
bit 0		CLC1IE: CLC1 Interrupt Enable bit									
		nterrupt enabled nterrupt disabled	1								
		nen upt uisabled	1								
Nete			must be								
	Bit PEIE of the IN set to enable any										

REGISTER 7-4: PIE3: PERIPHERAL INTERRUPT ENABLE REGISTER 3

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0				
_	TMR8IE	TMR5GIE	TMR5IE	TMR3GIE	TMR3IE	TMR6IE	TRM4IE				
bit 7							bit (
Legend:											
R = Readable	bit	W = Writable	bit	U = Unimplem	nented bit, read	as '0'					
u = Bit is uncł	nanged	x = Bit is unkr	iown	-n/n = Value a	t POR and BO	R/Value at all c	ther Resets				
'1' = Bit is set		'0' = Bit is clea	ared								
bit 7	Unimplemer	nted: Read as ')'								
bit 6	-			Enable bit							
	TMR8IE: TMR8 to T8PR Match Interrupt Enable bit 1 = Enables the Timer8 to T8PR match interrupt										
	0 = Disables the Timer8 to T8PR match interrupt										
bit 5	TMR5GIE: T	TMR5GIE: Timer5 Gate Interrupt Enable bit									
	1 = Enables	1 = Enables the Timer5 gate acquisition interrupt									
	0 = Disables	0 = Disables the Timer5 gate acquisition interrupt									
bit 4	TMR5IE: TMR5 to Overflow Interrupt Enable bit										
	1 = Enables the Timer5 to T5PR match interrupt										
	0 = Disables the Timer5 to T5PR match interrupt										
bit 3	TMR3GIE: Timer3 Gate Interrupt Enable bit										
	 1 = Enables the Timer3 gate acquisition interrupt 0 = Disables the Timer3 gate acquisition interrupt 										
bit 2		•	•								
	TMR3IE: TMR3 to Overflow Interrupt Enable bit										
	 Enables the Timer3 to T3PR match interrupt Disables the Timer3 to T3PR match interrupt 										
bit 1				•							
		TMR6IE: TMR6 to T6PR Match Interrupt Enable bit 1 = Enables the Timer6 to T6PR match interrupt									
		0 = Disables the Timer6 to T6PR match interrupt									
bit 0	TMR4IE: TM	R4 to T4PR Ma	tch Interrupt I	Enable bit							
	1 = Enables	the Timer4 to T4	4PR match in	terrupt							
	0 = Disables	the Timer4 to T	4PR match in	terrunt							

REGISTER 7-5: PIE4: PERIPHERAL INTERRUPT ENABLE REGISTER 4

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0				
CCP8IE ⁽¹⁾	CCP7IE	COG4IE	COG3IE	C8IE ⁽¹⁾	C7IE ⁽¹⁾	C6IE	C5IE				
bit 7							bit 0				
Legend: R = Readable	hit	W = Writable	hit.		nonted hit read						
		x = Bit is unki		•	nented bit, read at POR and BO		thar Baaata				
u = Bit is unch '1' = Bit is set	angeu	x = Bit is unki			al FOR and BO	R/Value at all 0	iller Resels				
I = Bit is set			areo								
bit 7	CCP8IF: CC	P8 Interrupt En	able bit ⁽¹⁾								
2	1 = Enables the CCP8 interrupt										
		s the CCP8 inte									
bit 6	CCP7IE: CC	P7 Interrupt En	able bit								
	1 = Enables the CCP7 interrupt										
		s the CCP7 inte	•								
bit 5		COG4IE: COG4 Auto-Shutdown Interrupt Enable bit									
		OG4 interrupt enabled									
L :1		0 = COG4 interrupt disabled									
bit 4	COG3IE: COG3 Auto-Shutdown Interrupt Enable bit										
	1 = COG3 interrupt enabled 0 = COG3 interrupt disabled										
bit 3	C8IE: Comparator C8 Interrupt Enable bit ⁽¹⁾										
	1 = Enables the Comparator C8 interrupt										
	0 = Disables the Comparator C8 interrupt										
bit 2	C7IE: Comparator C7 Interrupt Enable bit ⁽¹⁾										
	1 = Enables the Comparator C7 interrupt										
	0 = Disables the Comparator C7 interrupt										
bit 1	•	arator C6 Interr	•								
		the Comparato									
hit O		s the Comparate	•								
bit 0	•	arator C5 Interr the Comparato	•								
		s the Comparate									
	C16(L)F1777/9										
NOLE I. FIC	, io(L)F 1777/9	only.									

REGISTER 7-6: PIE5: PERIPHERAL INTERRUPT ENABLE REGISTER 5

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	
	—	_	_	PWM12IE ⁽¹⁾	PWM11IE	PWM6IE	PWM5IE	
bit 7							bit 0	
Legend:								
R = Readable	bit	W = Writable	bit	U = Unimplem	nented bit, read	as '0'		
u = Bit is unch	anged	x = Bit is unkr	iown	-n/n = Value a	t POR and BO	R/Value at all c	other Resets	
'1' = Bit is set		'0' = Bit is clea	ared					
bit 7-4	Unimplemen	ted: Read as ')'					
bit 3	PWM12IE: P	WM12 Interrupt	VM12 Interrupt Enable bit ⁽¹⁾					
		interrupt enable						
		interrupt disable						
bit 2		WM11 Interrupt						
		interrupt enable						
L :1		interrupt disable						
bit 1		VM6 Interrupt E nterrupt enabled						
		nterrupt disabled						
bit 0		VM5 Interrupt E						
		nterrupt enabled						
		nterrupt disable						
Note 1: PIC	C16(L)F1777/9	only.						
	. ,	2						

REGISTER 7-7: PIE6: PERIPHERAL INTERRUPT ENABLE REGISTER 6

R/W-0/0	R/W-0/0	R-0/0	R-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0			
TMR1GI	F ADIF	RCIF	TXIF	SSP1IF	CCP1IF	TMR2IF	TMR1IF			
bit 7			1		•		bit C			
Legend:										
R = Reada	ıble bit	W = Writable	bit	U = Unimplei	mented bit, read	d as '0'				
u = Bit is u	nchanged	x = Bit is unk	nown	-n/n = Value	at POR and BO	R/Value at all c	other Resets			
'1' = Bit is	set	'0' = Bit is cle	ared							
bit 7		Fimer1 Gate Inte	errupt Flag bit							
	1 = Interrup	t is pending t is not pending								
bit 6		og-to-Digital Cor	verter (ADC)	Interrunt Flag	ait					
bito	1 = Interrup			interrupt i lag i						
		t is not pending								
bit 5	RCIF: EUSA	ART Receive Int	errupt Flag bi	t						
	1 = Interrup									
	-	t is not pending								
bit 4		RT Transmit Inf	errupt Flag bi	t						
	1 = Interrup	t is penaing								
bit 3		nchronous Seria	al Port (MSSP) Interrupt Flag	bit					
	1 = Interrup			,						
		t is not pending								
bit 2	CCP1IF: CC	P1IF: CCP1 Interrupt Flag bit								
	1 = Interrup									
L	•	t is not pending	4	:4						
bit 1		ner2 to T2PR In	terrupt Flag b	DIT						
	1 = Interrup	t is not pending								
bit 0	-	mer1 Overflow li	nterrupt Flag I	oit						
	1 = Interrup									
	0 = Interrup	t is not pending								
Note:	Interrupt flag bits	are set when ar	n interrupt							
	condition occurs,	regardless of th	e state of							
	its corresponding									
	Enable bit, GIE, User software	should ens								
	appropriate inter	rrupt flag bits								
	prior to enabling									

REGISTER 7-8: PIR1: PERIPHERAL INTERRUPT REQUEST REGISTER 1

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R/W-0/0	R/W-0/0	R/W-0/0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0					
OSFIF	C2IF	C1IF	COG1IF	BCL1IF	C4IF	C3IF	CCP2IF					
bit 7	·						bit (
Legend:												
R = Reada		W = Writable		•	mented bit, read							
u = Bit is u	-	x = Bit is unk		-n/n = Value a	at POR and BO	R/Value at all c	other Resets					
'1' = Bit is	set	'0' = Bit is cle	ared									
bit 7	OSFIF: Osci	llator Fail Interr	upt Flag bit									
	1 = Interrupt	is pending										
	0 = Interrupt	is not pending										
bit 6	C2IF: Compa	arator C2 Interr	upt Flag bit									
	1 = Interrupt	is pending is not pending										
bit 5	•		unt Elag hit									
DIL J	•	C1IF: Comparator C1 Interrupt Flag bit 1 = Interrupt is pending										
		is not pending										
bit 4	COG1IF: CC)G1 Auto-Shuto	lown Interrupt	Flag bit								
	1 = Interrupt	1 = Interrupt is pending										
	0 = Interrupt	0 = Interrupt is not pending										
bit 3		BCL1IF: MSSP Bus Collision Interrupt Flag bit										
		1 = Interrupt is pending										
	-	0 = Interrupt is not pending										
bit 2		C4IF: Comparator C4 Interrupt Flag bit										
		1 = Interrupt is pending 0 = Interrupt is not pending										
bit 1	-	arator C3 Interr	upt Flag bit									
	1 = Interrupt		apt i lag bit									
		is not pending										
bit 0	CCP2IF: CC	P2 Interrupt Fla	ag bit									
	1 = Interrupt											
	0 = Interrupt	is not pending										
Note:	Interrupt flag bits											
	condition occurs, its corresponding											
	Enable bit, GIE,											
	User software	should ens	ure the									
	appropriate inter		are clear									
	prior to enabling a	an interrupt.										

REGISTER 7-9: PIR2: PERIPHERAL INTERRUPT REQUEST REGISTER 2

R/W-0/0) R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0					
—		COG2IF	ZCDIF	CLC4IF	CLC3IF	CLC2IF	CLC1IF					
bit 7	·						bit					
Legend:												
R = Reada		W = Writable			nented bit, read							
u = Bit is u		x = Bit is unkr		-n/n = Value a	at POR and BO	R/Value at all c	other Resets					
'1' = Bit is	set	'0' = Bit is cle	ared									
bit 7-6	Unimpleme	nted: Read as '	0'									
bit 5	COG2IF: CC	G2 Auto-Shutd	own Interrupt	Flag bit								
		1 = Interrupt is pending										
	•	is not pending										
bit 4		ZCDIF: Zero-Cross Detection Interrupt Flag bit										
		1 = Interrupt is pending 0 = Interrupt is not pending										
bit 3		C4 Interrupt Flag	n hit									
		= Interrupt is pending										
		Interrupt is not pending										
bit 2	CLC3IF: CL	_C3 Interrupt Flag bit										
		1 = Interrupt is pending										
	-	= Interrupt is not pending										
bit 1		CLC2IF: CLC2 Interrupt Flag bit										
		rrupt is pending rrupt is not pending										
bit 0		C1 Interrupt Flag	a bit									
	1 = Interrupt	-	5									
	0 = Interrupt	is not pending										
Note:	Interrupt flag bits condition occurs,											
	its corresponding											
	Enable bit, GIE,		•									
	User software appropriate inter	should ensu rupt flag bits a										
	prior to enabling a											
	. 0											

REGISTER 7-10: PIR3: PERIPHERAL INTERRUPT REQUEST REGISTER 3

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0				
—	TMR8IF	TMR5GIF	TMR5IF	TMR3GIF	TMR3IF	TMR6IF	TRM4IF				
bit 7							bit C				
Legend:											
R = Readable	bit	W = Writable	bit	U = Unimplem	nented bit, read	l as '0'					
u = Bit is unch	anged	x = Bit is unkr	iown	-n/n = Value a	t POR and BO	R/Value at all c	ther Resets				
'1' = Bit is set		'0' = Bit is clea	ared								
bit 7	Unimplemer	nted: Read as '	כי								
bit 6	-	R8 to T8PR Ma		-lag bit							
	1 = Interrupt is pending										
	0 = Interrupt	is not pending									
bit 5	TMR5GIF: Timer5 Gate Interrupt Flag bit										
	1 = Interrupt is pending										
	0 = Interrupt is not pending										
bit 4	TMR5IF: TMR5 to Overflow Interrupt Flag bit										
	1 = Interrupt is pending 0 = Interrupt is not pending										
bit 3	TMR3GIF: Timer3 Gate Interrupt Flag bit										
	1 = Interrupt is pending										
	0 = Interrupt is pending										
bit 2	TMR3IF: TM	R3 to Overflow	Interrupt Flag	bit							
	1 = Interrupt is pending										
	0 = Interrupt is not pending										
bit 1		R6 to T6PR Ma	tch Interrupt F	lag bit							
	1 = Interrupt is pending 0 = Interrupt is not pending										
L:1 0	-		4 - la 1 - 4 4 F								
bit 0	1 = Interrupt	R4 to T4PR Ma	ten interrupt i	-lag bit							
	0 = Interrupt										

REGISTER 7-11: PIR4: PERIPHERAL INTERRUPT REQUEST REGISTER 4

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0				
CCP8IF ⁽¹⁾	CCP7IF	COG4IF	COG3IF	C8IF ⁽¹⁾	C7IF ⁽¹⁾	C6IF	C5IF				
bit 7							bit				
Legend:											
R = Readable		W = Writable		•	mented bit, read						
u = Bit is unch	anged	x = Bit is unki		-n/n = Value a	at POR and BO	R/Value at all o	ther Resets				
'1' = Bit is set		'0' = Bit is cle	ared								
bit 7	CCP8IE: CC	P8 Interrunt Fla	a hit								
		CCP8IF: CCP8 Interrupt Flag bit 1 = Interrupt is pending									
		is not pending									
bit 6	CCP7IF: CCP7 Interrupt Flag bit										
	1 = Interrupt is pending 0 = Interrupt is not pending										
	-										
bit 5	COG4IF: COG4 Auto-Shutdown Interrupt Flag bit 1 = Interrupt is pending										
		is pending is not pending									
oit 4	COG3IF: COG3 Auto-Shutdown Interrupt Flag bit										
	1 = Interrupt is pending										
		is not pending									
bit 3	C8IF: Comparator C8 Interrupt Flag bit ⁽¹⁾										
	1 = Interrupt is pending										
1.11.0	 0 = Interrupt is not pending C7IF: Comparator C7 Interrupt Flag bit⁽¹⁾ 										
bit 2	•		upt Flag bit								
	1 = Interrupt is pending 0 = Interrupt is not pending										
bit 1	-	arator C6 Interri	upt Flag bit								
	1 = Interrupt										
	0 = Interrupt is not pending										
bit 0	•	arator C5 Interro	upt Flag bit								
	1 = Interrupt	is pending is not pending									
	v = merrupt	is nor denoind									

REGISTER 7-12: PIR5: PERIPHERAL INTERRUPT REQUEST REGISTER 5

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	
_	—	—	_	PWM12IF ⁽¹⁾	PWM11IF	PWM6IF	PWM5IF	
bit 7							bit 0	
Legend:								
R = Readable	e bit	W = Writable	bit	U = Unimplem	ented bit, read	as '0'		
u = Bit is unc	hanged	x = Bit is unkr	iown	-n/n = Value a	t POR and BOI	R/Value at all c	other Resets	
'1' = Bit is set		'0' = Bit is clea	ared					
bit 7-4	Unimplemen	ted: Read as ')'					
bit 3	PWM12IF: P	WM12 Interrupt Flag bit ⁽¹⁾						
	1 = Interrupt i	is pending	-					
	0 = Interrupt i	is not pending						
bit 2	PWM11IF: P\	PWM11IF: PWM11 Interrupt Flag bit						
	1 = Interrupt i							
		is not pending						
bit 1		/M6 Interrupt F	ag bit					
	1 = Interrupt i	is pending is not pending						
bit 0	•		aa hit					
DIEU	1 = Interrupt i	/M5 Interrupt Fl	ag bit					
		is not pending						
Note 1. Di	•							
Note 1: Pl	C16(L)F1777/9	orny.						

REGISTER 7-13: PIR6: PERIPHERAL INTERRUPT REQUEST REGISTER 6

						-		-	-
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	132
OPTION_REG	WPUEN	INTEDG	TMR0CS	TMR0SE	PSA		PS<2:0>		274
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	CCP1IE	TMR2IE	TMR1IE	133
PIE2	OSFIE	C2IE	C1IE	COG1IE	BCL1IE	C4IE	C3IE	CCP2IE	134
PIE3	—	_	COG2IE	ZCDIE	CLC4IE	CLC3IE	CLC2IE	CLC1IE	135
PIE4	_	TMR8IE	TMR5GIE	TMR5IE	TMR3GIE	TMR3IE	TMR6IE	TRM4IE	136
PIE5	CCP8IE ⁽¹⁾	CCP7IE	COG4IE ⁽¹⁾	COG3IE	C8IE ⁽¹⁾	C7IE ⁽¹⁾	C6IE	C5IE	137
PIE6	—	_	—	_	PWM12IE ⁽¹⁾	PWM11IE	PWM6IE	PWM5IE	138
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	CCP1IF	TMR2IF	TMR1IF	139
PIR2	OSFIF	C2IF	C1IF	COG1IF	BCL1IF	C4IF	C3IF	CCP2IF	140
PIR3	_	_	COG2IF	ZCDIF	CLC4IF	CLC3IF	CLC2IF	CLC1IF	141
PIR4	_	TMR8IF	TMR5GIF	TMR5IF	TMR3GIF	TMR3IF	TMR6IF	TRM4IF	142
PIR5	CCP8IF ⁽¹⁾	CCP7IF	COG4IF ⁽¹⁾	COG3IF	C8IF ⁽¹⁾	C7IF ⁽¹⁾	C6IF	C5IF	143
PIR6			—		PWM12IF ⁽¹⁾	PWM11IF	PWM6IF	PWM5IF	144

 TABLE 7-1:
 SUMMARY OF REGISTERS ASSOCIATED WITH INTERRUPTS

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by interrupts.

Note 1: PIC16(L)F1777/9 only.

8.0 POWER-DOWN MODE (SLEEP)

The Power-Down mode is entered by executing a SLEEP instruction.

Upon entering Sleep mode, the following conditions exist:

- 1. WDT will be cleared but keeps running, if enabled for operation during Sleep.
- 2. PD bit of the STATUS register is cleared.
- 3. TO bit of the STATUS register is set.
- 4. CPU clock is disabled.
- 5. 31 kHz LFINTOSC is unaffected and peripherals that operate from it may continue operation in Sleep.
- 6. Timer1 and peripherals that operate from Timer1 continue operation in Sleep when the Timer1 clock source selected is:
 - LFINTOSC
 - T1CKI
 - Secondary oscillator
- 7. ADC is unaffected, if the dedicated FRC oscillator is selected.
- I/O ports maintain the status they had before SLEEP was executed (driving high, low or high-impedance).
- 9. Resets other than WDT are not affected by Sleep mode.

Refer to individual chapters for more details on peripheral operation during Sleep.

To minimize current consumption, the following conditions should be considered:

- · I/O pins should not be floating
- External circuitry sinking current from I/O pins
- · Internal circuitry sourcing current from I/O pins
- Current draw from pins with internal weak pull-ups
- Modules using 31 kHz LFINTOSC
- Modules using secondary oscillator

I/O pins that are high-impedance inputs should be pulled to VDD or Vss externally to avoid switching currents caused by floating inputs.

Examples of internal circuitry that might be sourcing current include modules such as the DAC and FVR modules. See Section 17.0 "5-Bit Digital-to-Analog Converter (DAC) Module" and Section 14.0 "Fixed Voltage Reference (FVR)" for more information on these modules.

8.1 Wake-up from Sleep

The device can wake-up from Sleep through one of the following events:

- 1. External Reset input on MCLR pin, if enabled
- 2. BOR Reset, if enabled
- 3. POR Reset
- 4. Watchdog Timer, if enabled
- 5. Any external interrupt
- 6. Interrupts by peripherals capable of running during Sleep (see individual peripheral for more information)

The first three events will cause a device Reset. The last three events are considered a continuation of program execution. To determine whether a device Reset or wake-up event occurred, refer to Section 6.12 "Determining the Cause of a Reset".

When the SLEEP instruction is being executed, the next instruction (PC + 1) is prefetched. For the device to wake-up through an interrupt event, the corresponding interrupt enable bit must be enabled. Wake-up will occur regardless of the state of the GIE bit. If the GIE bit is disabled, the device continues execution at the instruction after the SLEEP instruction. If the GIE bit is enabled, the device executes the instruction after the SLEEP instruction, the device will then call the Interrupt Service Routine. In cases where the execution of the instruction following SLEEP is not desirable, the user should have a NOP after the SLEEP instruction.

The WDT is cleared when the device wakes up from Sleep, regardless of the source of wake-up.

8.1.1 WAKE-UP USING INTERRUPTS

When global interrupts are disabled (GIE cleared) and any interrupt source has both its interrupt enable bit and interrupt flag bit set, one of the following will occur:

- If the interrupt occurs **before** the execution of a SLEEP instruction
 - SLEEP instruction will execute as a NOP
 - WDT and WDT prescaler will not be cleared
 - TO bit of the STATUS register will not be set
 - PD bit of the STATUS register will not be cleared

- If the interrupt occurs **during or after** the execution of a SLEEP instruction
 - SLEEP instruction will be completely executed
 - Device will immediately wake-up from Sleep
 - WDT and WDT prescaler will be cleared
 - TO bit of the STATUS register will be set
 - PD bit of the STATUS register will be cleared

Even if the flag bits were checked before executing a SLEEP instruction, it may be possible for flag bits to become set before the SLEEP instruction completes. To determine whether a SLEEP instruction executed, test the PD bit. If the PD bit is set, the SLEEP instruction was executed as a NOP.

CLKIN ⁽¹ CLKOUT ⁽²	1 1 1	Q1 Q2 Q3 Q4 · Q1 Q2 Q3 Q4 · Q1 Q3 Q4 · Q1 Q2 Q3 Q4 · Q1 Q3 Q4 · Q1 Q2 Q3 Q4 · Q1 Q4 · Q1 ·
Interrupt flag		/ Interrupt Latency ⁽⁴⁾
GIE bit (INTCON reg) Processor ir Sleep	
Instruction Flow PC		C+2 PC+2 PC+2 0004h 0005h
Instruction J Fetched	Inst(PC) = Sleep Inst(PC + 1)	Inst(PC + 2) Inst(0004h) Inst(0005h)
Instruction J Executed	Inst(PC - 1) Sleep	Inst(PC + 1) Forced NOP Forced NOP Inst(0004h)
Note 1: 2: 3: 4:	"Two-Speed Clock Start-up Mode".	ed. EC, RC and INTOSC Oscillator modes or Two-Speed Start-up (see Section 5.4 e processor calls the ISR at 0004h. If GIE = 0, execution will continue in-line.

FIGURE 8-1: WAKE-UP FROM SLEEP THROUGH INTERRUPT

8.2 Low-Power Sleep Mode

The PIC16F1773/6 devices contain an internal Low Dropout (LDO) voltage regulator, which allows the device I/O pins to operate at voltages up to 5.5V while the internal device logic operates at a lower voltage. The LDO and its associated reference circuitry must remain active when the device is in Sleep mode. The PIC16F1773/6 allow the user to optimize the operating current in Sleep, depending on the application requirements.

A Low-Power Sleep mode can be selected by setting the VREGPM bit of the VREGCON register. With this bit set, the LDO and reference circuitry are placed in a low-power state when the device is in Sleep.

8.2.1 SLEEP CURRENT VS. WAKE-UP TIME

In the Default Operating mode, the LDO and reference circuitry remain in the normal configuration while in Sleep. The device is able to exit Sleep mode quickly since all circuits remain active. In Low-Power Sleep mode, when waking up from Sleep, an extra delay time is required for these circuits to return to the normal configuration and stabilize.

The Low-Power Sleep mode is beneficial for applications that stay in Sleep mode for long periods of time. The Normal mode is beneficial for applications that need to wake from Sleep quickly and frequently.

8.2.2 PERIPHERAL USAGE IN SLEEP

Some peripherals that can operate in Sleep mode will not operate properly with the Low-Power Sleep mode selected. The Low-Power Sleep mode is intended for use with the following peripherals only:

- Brown-out Reset (BOR)
- Watchdog Timer (WDT)
- · External interrupt pin/interrupt-on-change pins
- Timer1 (with external clock source < 100 kHz)
 - Note: The PIC16LF1777/8/9 do not have a configurable Low-Power Sleep mode. PIC16LF1777/8/9 are unregulated devices and are always in the lowest power state when in Sleep, with no wake-up time penalty. These devices have a lower maximum VDD and I/O voltage than the PIC16F1777/8/9. See Section 36.0 "Electrical Specifications" for more information.

8.3 Register Definitions: Voltage Regulator Control

U-0	U-0	U-0	U-0	U-0	U-0	R/W-0/0	R/W-1/1
	—	—	—	—	—	VREGPM	Reserved
bit 7							bit 0

REGISTER 8-1: VREGCON: VOLTAGE REGULATOR CONTROL REGISTER⁽¹⁾

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-2	Unimplemented: Read as '0'
bit 1	 VREGPM: Voltage Regulator Power Mode Selection bit 1 = Low-Power Sleep mode enabled in Sleep⁽²⁾ Draws lowest current in Sleep, slower wake-up 0 = Normal Power mode enabled in Sleep⁽²⁾ Draws higher current in Sleep, faster wake-up
bit 0	Reserved: Read as '1'. Maintain this bit set.

Note 1: PIC16F1777/8/9 only.

2: See Section 36.0 "Electrical Specifications".

PIC16(L)F1777/8/9

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	132
IOCAP	IOCAP7	IOCAP6	IOCAP5	IOCAP4	IOCAP3	IOCAP2	IOCAP1	IOCAP0	215
IOCAN	IOCAN7	IOCAN6	IOCAN5	IOCAN4	IOCAN3	IOCAN2	IOCAN1	IOCAN0	215
IOCAF	IOCAF7	IOCAF6	IOCAF5	IOCAF4	IOCAF3	IOCAF2	IOCAF1	IOCAF0	215
IOCBP	IOCBP7	IOCBP6	IOCBP5	IOCBP4	IOCBP3	IOCBP2	IOCBP1	IOCBP0	216
IOCBN	IOCBN7	IOCBN6	IOCBN5	IOCBN4	IOCBN3	IOCBN2	IOCBN1	IOCBN0	216
IOCBF	IOCBF7	IOCBF6	IOCBF5	IOCBF4	IOCBF3	IOCBF2	IOCBF1	IOCBF0	216
IOCCP	IOCCP7	IOCCP6	IOCCP5	IOCCP4	IOCCP3	IOCCP2	IOCCP1	IOCCP0	216
IOCCN	IOCCN7	IOCCN6	IOCCN5	IOCCN4	IOCCN3	IOCCN2	IOCCN1	IOCCN0	216
IOCCF	IOCCF7	IOCCF6	IOCCF5	IOCCF4	IOCCF3	IOCCF2	IOCCF1	IOCCF0	216
IOCEP	_	_	_	_	IOCEP3	—	_	_	219
IOCEN			—		IOCEN3	—	—		219
IOCEF	_	_	_	_	IOCEF3	_	_	_	220
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	CCP1IE	TMR2IE	TMR1IE	133
PIE2	OSFIE	C2IE	C1IE	COG1IE	BCL1IE	C4IE	C3IE	CCP2IE	134
PIE3		_	COG2IE	ZCDIE	CLC4IE	CLC3IE	CLC2IE	CLC1IE	135
PIE4	—	TMR8IE	TMR5GIE	TMR5IE	TMR3GIE	TMR3IE	TMR6IE	TRM4IE	136
PIE5	CCP8IE ⁽¹⁾	CCP7IE	COG4IE ⁽¹⁾	COG3IE	C8IE ⁽¹⁾	C7IE ⁽¹⁾	C6IE	C5IE	137
PIE6		_			PWM12IE ⁽¹⁾	PWM11IE	PWM6IE	PWM5IE	138
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	CCP1IF	TMR2IF	TMR1IF	139
PIR2	OSFIF	C2IF	C1IF	COG1IF	BCL1IF	C4IF	C3IF	CCP2IF	140
PIR3	_		COG2IF	ZCDIF	CLC4IF	CLC3IF	CLC2IF	CLC1IF	141
PIR4	_	TMR8IF	TMR5GIF	TMR5IF	TMR3GIF	TMR3IF	TMR6IF	TRM4IF	142
PIR5	CCP8IF ⁽¹⁾	CCP7IF	COG4IF ⁽¹⁾	COG3IF	C8IF ⁽¹⁾	C7IF ⁽¹⁾	C6IF	C5IF	143
PIR6	—	—	—	—	PWM12IF ⁽¹⁾	PWM11IF	PWM6IF	PWM5IF	144
STATUS	_	_	—	TO	PD	Z	DC	С	40
VREGCON ⁽¹⁾	_		—	_	—	—	VREGPM ⁽²⁾	Reserved	149
WDTCON	_	—			WDTPS<4:0>			SWDTEN	154

TABLE 8-1: SUMMARY OF REGISTERS ASSOCIATED WITH POWER-DOWN MODE

Legend: — = unimplemented location, read as '0'. Shaded cells are not used in Power-Down mode.

Note 1: PIC16(L)F1777/9 only.

2: Unimplemented on PIC16LF1777/8/9.

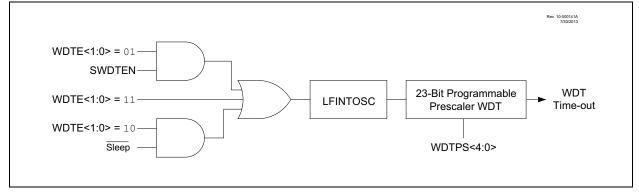
9.0 WATCHDOG TIMER (WDT)

The Watchdog Timer is a system timer that generates a Reset if the firmware does not issue a CLRWDT instruction within the time-out period. The Watchdog Timer is typically used to recover the system from unexpected events.

The WDT has the following features:

- · Independent clock source
- Multiple operating modes
 - WDT is always on
 - WDT is off when in Sleep
 - WDT is controlled by software
 - WDT is always off
- Configurable time-out period is from 1 ms to 256 seconds (nominal)
- Multiple Reset conditions
- Operation during Sleep

FIGURE 9-1: WATCHDOG TIMER BLOCK DIAGRAM



9.1 Independent Clock Source

The WDT derives its time base from the 31 kHz LFINTOSC internal oscillator. Time intervals in this chapter are based on a nominal interval of 1 ms. See Table 36-8: Oscillator Parameters for the LFINTOSC specification.

9.2 WDT Operating Modes

The Watchdog Timer module has four operating modes controlled by the WDTE<1:0> bits in Configuration Words. See Table 9-1.

9.2.1 WDT IS ALWAYS ON

When the WDTE bits of Configuration Words are set to '11', the WDT is always on.

WDT protection is active during Sleep.

9.2.2 WDT IS OFF IN SLEEP

When the WDTE bits of Configuration Words are set to '10', the WDT is on, except in Sleep.

WDT protection is not active during Sleep.

9.2.3 WDT CONTROLLED BY SOFTWARE

When the WDTE bits of Configuration Words are set to '01', the WDT is controlled by the SWDTEN bit of the WDTCON register.

WDT protection is unchanged by Sleep. See Table 9-1 for more details.

WDTE<1:0>	SWDTEN	Device Mode	WDT Mode
11	Х	Х	Active
10	37	Awake	Active
TO	X Sleep		Disabled
01	1	х	Active
UT	0	~	Disabled
00	Х	х	Disabled

TABLE 9-1: WDT OPERATING MODES

9.3 Time-Out Period

The WDTPS bits of the WDTCON register set the time-out period from 1 ms to 256 seconds (nominal). After a Reset, the default time-out period is two seconds.

9.4 Clearing the WDT

The WDT is cleared when any of the following conditions occur:

- Any Reset
- CLRWDT instruction is executed
- Device enters Sleep
- · Device wakes up from Sleep
- · Oscillator fail
- · WDT is disabled
- Oscillator Start-up Timer (OST) is running

See Table 9-2 for more information.

9.5 Operation During Sleep

When the device enters Sleep, the WDT is cleared. If the WDT is enabled during Sleep, the WDT resumes counting.

When the device exits Sleep, the WDT is cleared again. The WDT remains clear until the OST, if enabled, completes. See **Section 5.0** "Oscillator **Module (with Fail-Safe Clock Monitor)**" for more information on the OST.

When a WDT time-out occurs while the device is in Sleep, no Reset is generated. Instead, the device wakes up and resumes operation. The TO and PD bits in the STATUS register are changed to indicate the event. See STATUS Register (Register 3-1) for more information.

TABLE 9-2: WDT CLEARING CONDITIONS

Conditions	WDT			
WDTE<1:0> = 00				
WDTE<1:0> = 01 and SWDTEN = 0				
WDTE<1:0> = 10 and enter Sleep	Cleared			
CLRWDT Command	Cleared			
Oscillator Fail Detected				
Exit Sleep + System Clock = T1OSC, EXTRC, INTOSC, EXTCLK				
Exit Sleep + System Clock = XT, HS, LP	Cleared until the end of OST			
Change INTOSC divider (IRCF bits)	Unaffected			

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9.6 Register Definitions: Watchdog Control

REGISTER 9-1: WDTCON: WATCHDOG TIMER CONTROL REGISTER

U-0	U-0	R/W-0/0	R/W-1/1	R/W-0/0	R/W-1/1	R/W-1/1	R/W-0/0
_	—			WDTPS<4:0>	(1)		SWDTEN
bit 7		·					bit 0
Legend:							
R = Readable	e bit	W = Writable	bit	U = Unimpler	mented bit, read	l as '0'	
u = Bit is unc	hanged	x = Bit is unki	nown	-m/n = Value	at POR and BC	R/Value at all	other Resets
'1' = Bit is se	t	'0' = Bit is cle	ared				
bit 7-6	Unimplom	ented: Read as '	0'				
bit 5-1	-	0>: Watchdog Ti		elect hits(1)			
		Prescale Rate					
			s in minimum	intonyal (1.32)			
	•	Reserved. Result	s in minimum				
	•						
	•						
	10011 = R	Reserved. Result	s in minimum	interval (1:32)			
	10010 = 1	:8388608 (2 ²³) (Interval 256s	nominal)			
		:4194304 (2 ²²) (
	10000 = 1	:2097152 (2 ²¹) (Interval 64s n	ominal)			
	01111 = 1	:1048576 (2 ²⁰) (Interval 32s n	ominal)			
	01110 = 1	:524288 (2 ¹⁹) (Ir	nterval 16s no	minal)			
	01101 = 1	:262144 (2 ¹⁸) (Ir :131072 (2 ¹⁷) (Ir	nterval 8s non	ninal)			
		:65536 (Interval					
		:32768 (Interval					
		:16384 (Interval		nal)			
		:8192 (Interval 2		,			
		:4096 (Interval 1					
	00110 = 1	:2048 (Interval 6	4 ms nominal)			
		:1024 (Interval 3)			
		:512 (Interval 16	,				
		:256 (Interval 8)					
		:128 (Interval 4 i :64 (Interval 2 m					
		:32 (Interval 1 m	,				
bit 0		Software Enable	-	/atchdog Timer	bit		
	If WDTE<1:			atonaog milei	Sit		
	This bit is ig						
	If WDTE<1:						
	1 = WDT is						
	0 = WDT is						
	<u>If WDTE<1:</u>						
	This bit is ig	nored.					



Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
OSCCON	SPLLEN		IRCF	<3:0>		—	SCS	<1:0>	116
STATUS	—	—	—	TO	PD	Z	DC	С	40
WDTCON	—	—		١	WDTPS<4:0		SWDTEN	154	

TABLE 9-3: SUMMARY OF REGISTERS ASSOCIATED WITH WATCHDOG TIMER

Legend: x = unknown, u = unchanged, – = unimplemented locations read as '0'. Shaded cells are not used by Watchdog Timer.

TABLE 9-4:	SUMMARY OF CONFIGURATION WORD WITH WATCHDOG TIMER
IADLL J-T.	

Name	Bits	Bit -/7	Bit -/6	Bit 13/5	Bit 12/4	Bit 11/3	Bit 10/2	Bit 9/1	Bit 8/0	Register on Page
CONFIG1	13:8	_	_	FCMEN	IESO	CLKOUTEN	BOREN<1:0>			95
CONFIGT	7:0	CP	CP MCLRE PWRTE WDTE<1:0>		E<1:0>	F	OSC<2:0	>	90	

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by Watchdog Timer.

10.0 FLASH PROGRAM MEMORY CONTROL

The Flash program memory is readable and writable during normal operation over the full VDD range. Program memory is indirectly addressed using Special Function Registers (SFRs). The SFRs used to access program memory are:

- PMCON1
- PMCON2
- PMDATL
- PMDATH
- PMADRL
- PMADRH

When accessing the program memory, the PMDATH:PMDATL register pair forms a 2-byte word that holds the 14-bit data for read/write, and the PMADRH:PMADRL register pair forms a 2-byte word that holds the 15-bit address of the program memory location being read.

The write time is controlled by an on-chip timer. The write/erase voltages are generated by an on-chip charge pump rated to operate over the operating voltage range of the device.

The Flash program memory can be protected in two ways; by code protection (CP bit in Configuration Words) and write protection (WRT<1:0> bits in Configuration Words).

Code protection $(\overline{CP} = 0)^{(1)}$, disables access, reading and writing, to the Flash program memory via external device programmers. Code protection does not affect the self-write and erase functionality. Code protection can only be reset by a device programmer performing a Bulk Erase to the device, clearing all Flash program memory, Configuration bits and User IDs.

Write protection prohibits self-write and erase to a portion or all of the Flash program memory as defined by the bits, WRT<1:0>. Write protection does not affect a device programmers ability to read, write or erase the device.

Note 1:	Code	protection	of	the	entire	Flash
	progra	m m <u>em</u> ory	ar	ray i	s enab	led by
	clearin	g the CP bit	of C	Config	uration	Words.

10.1 PMADRL and PMADRH Registers

The PMADRH:PMADRL register pair can address up to a maximum of 32K words of program memory. When selecting a program address value, the MSB of the address is written to the PMADRH register and the LSB is written to the PMADRL register.

10.1.1 PMCON1 AND PMCON2 REGISTERS

PMCON1 is the control register for Flash program memory accesses.

Control bits RD and WR initiate read and write, respectively. These bits cannot be cleared, only set, in software. They are cleared by hardware at completion of the read or write operation. The inability to clear the WR bit in software prevents the accidental, premature termination of a write operation.

The WREN bit, when set, will allow a write operation to occur. On power-up, the WREN bit is clear. The WRERR bit is set when a write operation is interrupted by a Reset during normal operation. In these situations, following Reset, the user can check the WRERR bit and execute the appropriate error handling routine.

The PMCON2 register is a write-only register. Attempting to read the PMCON2 register will return all '0's.

To enable writes to the program memory, a specific pattern (the unlock sequence), must be written to the PMCON2 register. The required unlock sequence prevents inadvertent writes to the program memory write latches and Flash program memory.

10.2 Flash Program Memory Overview

It is important to understand the Flash program memory structure for erase and programming operations. Flash program memory is arranged in rows. A row consists of a fixed number of 14-bit program memory words. A row is the minimum size that can be erased by user software.

After a row has been erased, the user can reprogram all or a portion of this row. Data to be written into the program memory row is written to 14-bit wide data write latches. These write latches are not directly accessible to the user, but may be loaded via sequential writes to the PMDATH:PMDATL register pair.

See Table 10-1 for Erase Row size and the number of write latches for Flash program memory.

Note: If the user wants to modify only a portion of a previously programmed row, then the contents of the entire row must be read and saved in RAM prior to the erase. Then, new data and retained data can be written into the write latches to reprogram the row of Flash program memory. However, any unprogrammed locations can be written without first erasing the row. In this case, it is not necessary to save and rewrite the other previously programmed locations.

TABLE 10-1:FLASH MEMORY
ORGANIZATION BY DEVICE

Device	Row Erase (words)	Write Latches (words)	
PIC16(L)F1778	32	32	
PIC16(L)F1777/9	52	52	

10.2.1 READING THE FLASH PROGRAM MEMORY

To read a program memory location, the user must:

- 1. Write the desired address to the PMADRH:PMADRL register pair.
- 2. Clear the CFGS bit of the PMCON1 register.
- 3. Then, set control bit RD of the PMCON1 register.

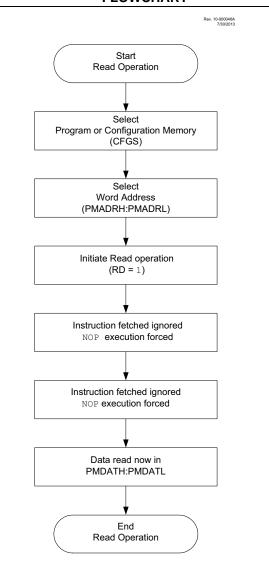
Once the read control bit is set, the program memory Flash controller will use the second instruction cycle to read the data. This causes the second instruction immediately following the "BSF PMCON1, RD" instruction to be ignored. The data is available in the very next cycle, in the PMDATH:PMDATL register pair; therefore, it can be read as two bytes in the following instructions.

The PMDATH:PMDATL register pair will hold this value until another read or until it is written to by the user.

Note:	The two instructions following a program			
	memory read are required to be NOPS.			
	This prevents the user from executing a			
	2-cycle instruction on the next instruction			
	after the RD bit is set.			

FIGURE 10-1:

FLASH PROGRAM MEMORY READ FLOWCHART



PIC16(L)F1777/8/9

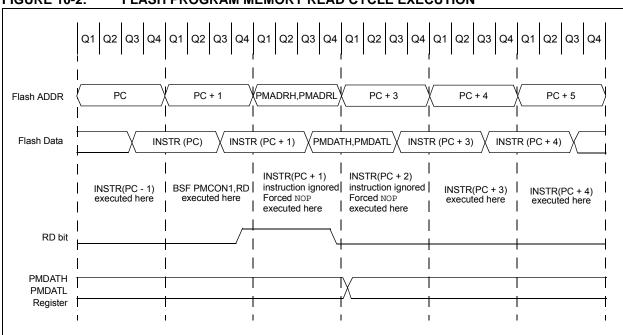


FIGURE 10-2: FLASH PROGRAM MEMORY READ CYCLE EXECUTION

EXAMPLE 10-1: FLASH PROGRAM MEMORY READ

```
* This code block will read 1 word of program
```

- * memory at the memory address:
- PROG_ADDR_HI : PROG_ADDR_LO
- * data will be returned in the variables;
- * PROG_DATA_HI, PROG_DATA_LO

BANKSEL	PMADRL	; Select Bank for PMCON registers
MOVLW	PROG_ADDR_LO	;
MOVWF	PMADRL	; Store LSB of address
MOVLW	PROG_ADDR_HI	;
MOVWF	PMADRH	; Store MSB of address
BCF BSF NOP NOP	PMCON1,CFGS PMCON1,RD	<pre>; Do not select Configuration Space ; Initiate read ; Ignored (Figure 10-1) ; Ignored (Figure 10-1)</pre>
MOVF	PMDATL,W	; Get LSB of word
MOVWF	PROG_DATA_LO	; Store in user location
MOVF	PMDATH,W	; Get MSB of word
MOVWF	PROG_DATA_HI	; Store in user location

10.2.2 FLASH MEMORY UNLOCK SEQUENCE

The unlock sequence is a mechanism that protects the Flash program memory from unintended self-write programming or erasing. The sequence must be executed and completed without interruption to successfully complete any of the following operations:

- Row Erase
- · Load program memory write latches
- Write of program memory write latches to program memory
- Write of program memory write latches to User IDs

The unlock sequence consists of the following steps:

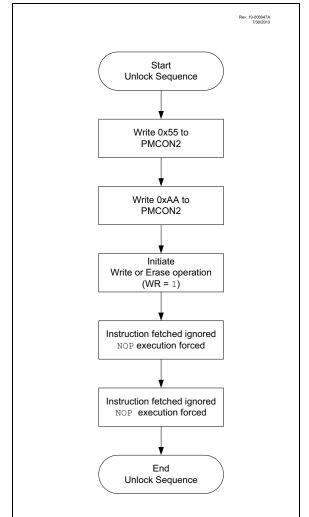
- 1. Write 55h to PMCON2
- 2. Write AAh to PMCON2
- 3. Set the WR bit in PMCON1
- 4. NOP instruction
- 5. NOP instruction

Once the WR bit is set, the processor will always force two NOP instructions. When an Erase Row or Program Row operation is being performed, the processor will stall internal operations (typical 2 ms), until the operation is complete and then resume with the next instruction. When the operation is loading the program memory write latches, the processor will always force the two NOP instructions and continue uninterrupted with the next instruction.

Since the unlock sequence must not be interrupted, global interrupts should be disabled prior to the unlock sequence and re-enabled after the unlock sequence is completed.

FIGURE 10-3:

FLASH PROGRAM MEMORY UNLOCK SEQUENCE FLOWCHART



PIC16(L)F1777/8/9

10.2.3 ERASING FLASH PROGRAM MEMORY

While executing code, program memory can only be erased by rows. To erase a row:

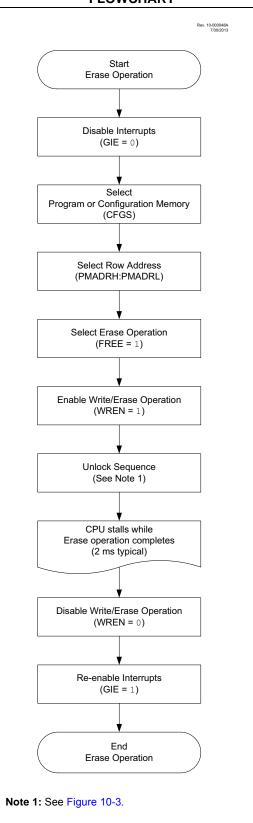
- 1. Load the PMADRH:PMADRL register pair with any address within the row to be erased.
- 2. Clear the CFGS bit of the PMCON1 register.
- 3. Set the FREE and WREN bits of the PMCON1 register.
- 4. Write 55h, then AAh, to PMCON2 (Flash programming unlock sequence).
- 5. Set control bit WR of the PMCON1 register to begin the erase operation.

See Example 10-2.

After the "BSF PMCON1, WR" instruction, the processor requires two cycles to set up the erase operation. The user must place two NOP instructions immediately following the WR bit set instruction. The processor will halt internal operations for the typical 2 ms erase time. This is not Sleep mode as the clocks and peripherals will continue to run. After the erase cycle, the processor will resume operation with the third instruction after the PMCON1 write instruction.

FIGURE 10-4:

FLASH PROGRAM MEMORY ERASE FLOWCHART



PIC16(L)F1777/8/9

EXAMPLE 10-2: ERASING ONE ROW OF PROGRAM MEMORY

- ; This row erase routine assumes the following:
- ; 1. A valid address within the erase row is loaded in ADDRH:ADDRL

; 2. ADDRH and ADDRL are located in shared data memory $0\,\mathrm{x}70$ - $0\,\mathrm{x}7F$ (common RAM)

	BCF BANKSEL MOVF MOVF MOVF BCF BCF BSF BSF	INTCON, GIE PMADRL ADDRL,W PMADRL ADDRH,W PMADRH PMCON1,CFGS PMCON1,FREE PMCON1,WREN	<pre>; Disable ints so required sequences will execute properly ; Load lower 8 bits of erase address boundary ; Load upper 6 bits of erase address boundary ; Not configuration space ; Specify an erase operation ; Enable writes</pre>
Required Sequence	MOVLW MOVUF MOVLW MOVWF BSF NOP NOP	55h PMCON2 0AAh PMCON2 PMCON1,WR	<pre>; Start of required sequence to initiate erase ; Write 55h ; ; Write AAh ; Set WR bit to begin erase ; NOP instructions are forced as processor starts ; row erase of program memory. ; ; The processor stalls until the erase process is complete ; after erase processor continues with 3rd instruction</pre>
	BCF BSF	PMCON1,WREN INTCON,GIE	; Disable writes ; Enable interrupts

10.2.4 WRITING TO FLASH PROGRAM MEMORY

Program memory is programmed using the following steps:

- 1. Load the address in PMADRH:PMADRL of the row to be programmed.
- 2. Load each write latch with data.
- 3. Initiate a programming operation.
- 4. Repeat steps 1 through 3 until all data is written.

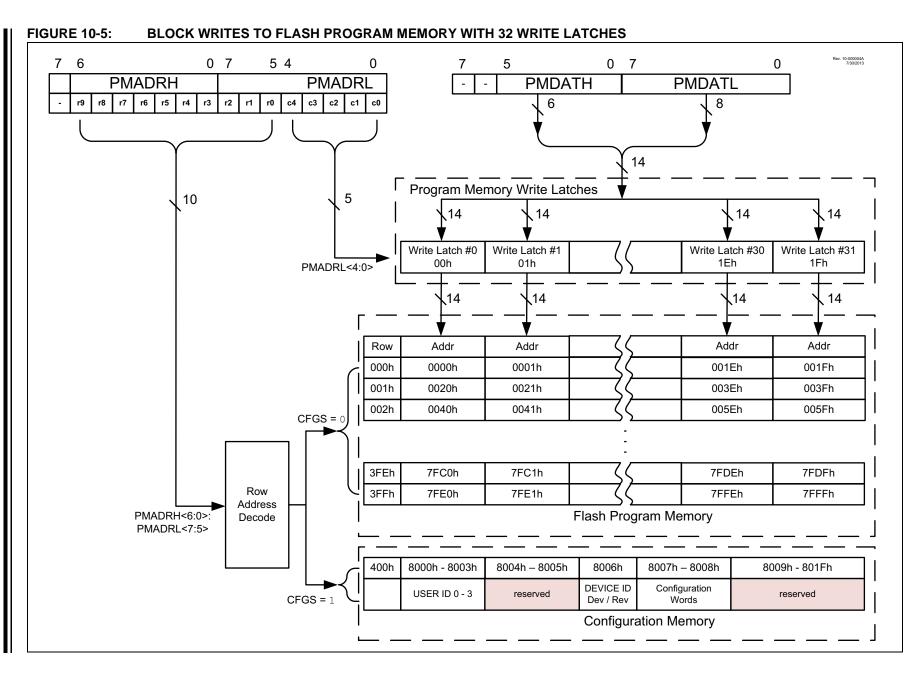
Before writing to program memory, the word(s) to be written must be erased or previously unwritten. Program memory can only be erased one row at a time. No automatic erase occurs upon the initiation of the write.

Program memory can be written one or more words at a time. The maximum number of words written at one time is equal to the number of write latches. See Figure 10-5 (row writes to program memory with 32 write latches) for more details.

The write latches are aligned to the Flash row address boundary defined by the upper ten bits of PMADRH:PMADRL, (PMADRH<6:0>:PMADRL<7:5>) with the lower five bits of PMADRL, (PMADRL<7:5>) determining the write latch being loaded. Write operations do not cross these boundaries. At the completion of a program memory write operation, the data in the write latches is reset to contain 0x3FFF. The following steps should be completed to load the write latches and program a row of program memory. These steps are divided into two parts. First, each write latch is loaded with data from the PMDATH:PMDATL using the unlock sequence with LWLO = 1. When the last word to be loaded into the write latch is ready, the LWLO bit is cleared and the unlock sequence executed. This initiates the programming operation, writing all the latches into Flash program memory.

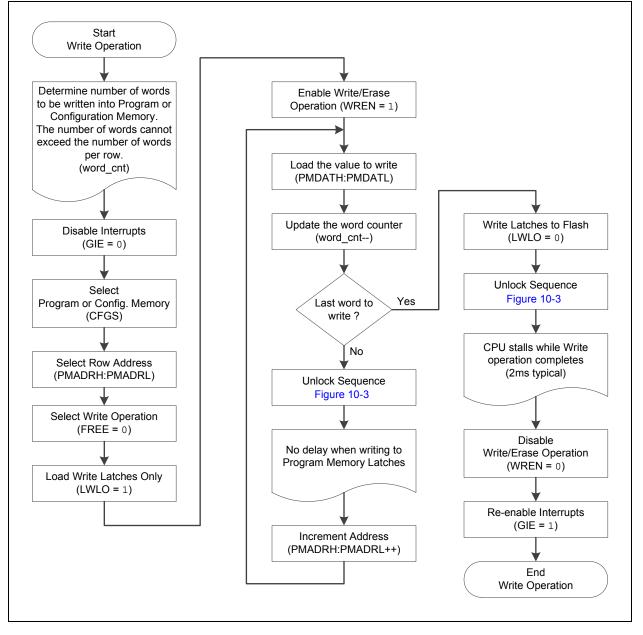
- Note: The special unlock sequence is required to load a write latch with data or initiate a Flash programming operation. If the unlock sequence is interrupted, writing to the latches or program memory will not be initiated.
- 1. Set the WREN bit of the PMCON1 register.
- 2. Clear the CFGS bit of the PMCON1 register.
- Set the LWLO bit of the PMCON1 register. When the LWLO bit of the PMCON1 register is '1', the write sequence will only load the write latches and will not initiate the write to Flash program memory.
- 4. Load the PMADRH:PMADRL register pair with the address of the location to be written.
- 5. Load the PMDATH:PMDATL register pair with the program memory data to be written.
- Execute the unlock sequence (Section 10.2.2 "Flash Memory Unlock Sequence"). The write latch is now loaded.
- 7. Increment the PMADRH:PMADRL register pair to point to the next location.
- 8. Repeat steps 5 through 7 until all but the last write latch has been loaded.
- Clear the LWLO bit of the PMCON1 register. When the LWLO bit of the PMCON1 register is '0', the write sequence will initiate the write to Flash program memory.
- 10. Load the PMDATH:PMDATL register pair with the program memory data to be written.
- 11. Execute the unlock sequence (Section 10.2.2 "Flash Memory Unlock Sequence"). The entire program memory latch content is now written to Flash program memory.
 - Note: The program memory write latches are reset to the blank state (0x3FFF) at the completion of every write or erase operation. As a result, it is not necessary to load all the program memory write latches. Unloaded latches will remain in the blank state.

An example of the complete write sequence is shown in Example 10-3. The initial address is loaded into the PMADRH:PMADRL register pair; the data is loaded using indirect addressing.



IC16(L)F1777/8/9

FIGURE 10-6: FLASH PROGRAM MEMORY WRITE FLOWCHART



EXAMPLE 10-3: WRITING TO FLASH PROGRAM MEMORY

; This write routine assumes the following: ; 1. 64 bytes of data are loaded, starting at the address in DATA_ADDR ; 2. Each word of data to be written is made up of two adjacent bytes in DATA_ADDR, ; stored in little endian format ; 3. A valid starting address (the least significant bits = 00000) is loaded in ADDRH: ADDRL ; 4. ADDRH and ADDRL are located in shared data memory 0x70 - 0x7F (common RAM) ; BCF INTCON,GIE ; Disable ints so required sequences will execute properly ; Bank 3 BANKSEL PMADRH MOVF ADDRH,W ; Load initial address MOVWF PMADRH MOVF ADDRL,W MOVWF PMADRL LOW DATA_ADDR ; Load initial data address MOVLW MOVWF FSROL ; MOVLW HIGH DATA_ADDR ; Load initial data address MOVWF FSR0H ; PMCON1,CFGS ; Not configuration space BCF BSF PMCON1,WREN ; Enable writes PMCON1,LWLO ; Only Load Write Latches BSF LOOP MOVIW FSR0++ ; Load first data byte into lower MOVWF PMDATT. ; ; Load second data byte into upper MOVIW FSR0++ MOVWF PMDATH MOVF ; Check if lower bits of address are '00000' PMADRL,W ; Check if we're on the last of 32 addresses XORLW 0x1F ANDLW 0x1F BTFSC STATUS,Z ; Exit if last of 32 words, GOTO START_WRITE ; MOVLW 55h ; Start of required write sequence: MOVWF PMCON2 ; Write 55h Required Sequence MOVLW 0AAh MOVWF PMCON2 ; Write AAh BSF ; Set WR bit to begin write PMCON1,WR NOP ; NOP instructions are forced as processor ; loads program memory write latches NOP INCF PMADRL, F ; Still loading latches Increment address GOTO LOOP ; Write next latches START_WRITE BCF PMCON1,LWLO ; No more loading latches - Actually start Flash program ; memory write MOVLW 55h ; Start of required write sequence: MOVWF PMCON2 ; Write 55h Required Sequence MOVLW 0AAh ; MOVWF PMCON2 ; Write AAh BSF PMCON1,WR ; Set WR bit to begin write NOP ; NOP instructions are forced as processor writes ; all the program memory write latches simultaneously NOP ; to program memory. ; After NOPs, the processor ; stalls until the self-write process in complete ; after write processor continues with 3rd instruction PMCON1,WREN BCF ; Disable writes BSF INTCON,GIE ; Enable interrupts

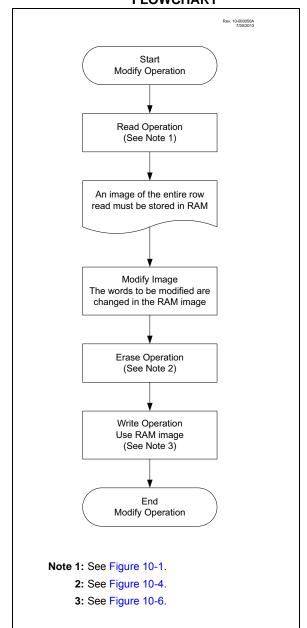
10.3 Modifying Flash Program Memory

When modifying existing data in a program memory row, and data within that row must be preserved, it must first be read and saved in a RAM image. Program memory is modified using the following steps:

- 1. Load the starting address of the row to be modified.
- 2. Read the existing data from the row into a RAM image.
- 3. Modify the RAM image to contain the new data to be written into program memory.
- 4. Load the starting address of the row to be rewritten.
- 5. Erase the program memory row.
- 6. Load the write latches with data from the RAM image.
- 7. Initiate a programming operation.

FIGURE 10-7:

FLASH PROGRAM MEMORY MODIFY FLOWCHART



10.4 User ID, Device ID and Configuration Word Access

Instead of accessing program memory, the User ID's, Device ID/Revision ID and Configuration Words can be accessed when CFGS = 1 in the PMCON1 register. This is the region that would be pointed to by PC<15> = 1, but not all addresses are accessible. Different access may exist for reads and writes. Refer to Table 10-2.

When read access is initiated on an address outside the parameters listed in Table 10-2, the PMDATH:PMDATL register pair is cleared, reading back '0's.

Address	Function	Read Access	Write Access
8000h-8003h	User IDs	Yes	Yes
8005h-8006h	Device ID/Revision ID	Yes	No
8007h-8008h	Configuration Words 1 and 2	Yes	No

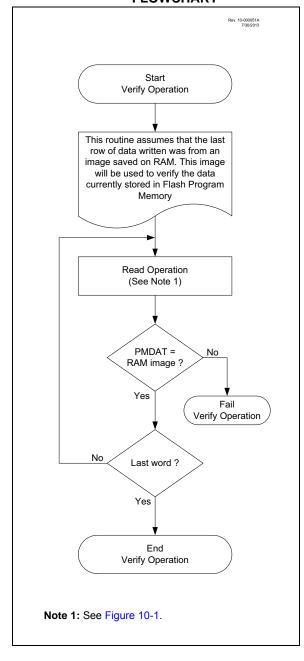
EXAMPLE 10-4: CONFIGURATION WORD AND DEVICE ID ACCESS

* PROG_ADDI		word of program memory at the memory address: -08h) data will be returned in the variables;)		
BANKSEL	PMADRL	; Select correct Bank		
MOVLW	PROG_ADDR_LO	;		
MOVWF	PMADRL	; Store LSB of address		
CLRF	PMADRH	; Clear MSB of address		
BSF	PMCON1,CFGS	; Select Configuration Space		
BCF	INTCON,GIE	; Disable interrupts		
BSF	BSF PMCON1,RD ; Initiate read			
NOP		; Executed (See Figure 10-2)		
NOP		; Ignored (See Figure 10-2)		
BSF	INTCON,GIE	; Restore interrupts		
MOVF	PMDATL,W	; Get LSB of word		
MOVWF	PROG_DATA_LO	; Store in user location		
MOVF	PMDATH,W	; Get MSB of word		
MOVWF	PROG_DATA_HI	; Store in user location		

10.5 Write/Verify

It is considered good programming practice to verify that program memory writes agree with the intended value. Since program memory is stored as a full page then the stored program memory contents are compared with the intended data stored in RAM after the last write is complete.

FIGURE 10-8: FLASH PROGRAM MEMORY VERIFY FLOWCHART



10.6 Register Definitions: Flash Program Memory Control

REGISTER 10-1: PMDATL: PROGRAM MEMORY DATA LOW BYTE REGISTER

	D 4 4 4						
R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
			PMD	AT<7:0>			
bit 7							bit 0
Legend:							
R = Readable I	bit	W = Writable b	oit	U = Unimpler	nented bit, read	l as '0'	
u = Bit is uncha	anged	x = Bit is unkn	own	-n/n = Value a	at POR and BO	R/Value at all o	other Resets
'1' = Bit is set		'0' = Bit is clea	ared				

bit 7-0 **PMDAT<7:0>**: Read/write value for Least Significant bits of program memory

REGISTER 10-2: PMDATH: PROGRAM MEMORY DATA HIGH BYTE REGISTER

U-0	U-0	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
—	—			PMDA	T<13:8>		
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-6	Unimplemented: Read as '0'
---------	----------------------------

bit 5-0 **PMDAT<13:8>**: Read/write value for Most Significant bits of program memory

REGISTER 10-3: PMADRL: PROGRAM MEMORY ADDRESS LOW BYTE REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	
PMADR<7:0>								
bit 7								

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 PMADR<7:0>: Specifies the Least Significant bits for program memory address

U-1	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
(1)				PMADR<14:8	>		
bit 7							bit 0
Legend:							
R = Readable b	oit	W = Writable I	bit	U = Unimpler	nented bit, read	l as '0'	
u = Bit is unchanged x = Bit is unknown -n/n = Value at POR and BOR/Value at all othe				other Resets			
'1' = Bit is set		'0' = Bit is clea	ared				

bit 7 Unimplemented: Read as '1'

bit 6-0 PMADR<14:8>: Specifies the Most Significant bits for program memory address

Note 1: Unimplemented, read as '1'.

U-1	R/W-0/0	R/W-0/0	R/W/HC-0/0	R/W/HC-x/q ⁽²⁾	R/W-0/0	R/S/HC-0/0	R/S/HC-0/0
_(1)	CFGS	LWLO ⁽³⁾	FREE	WRERR	WREN	WR	RD
bit 7				· ·			bit 0
Legend:							
R = Readabl	e bit	W = Writable	bit	U = Unimplemer			
S = Bit can o	5	x = Bit is unkn		-n/n = Value at P	OR and BOR	Value at all othe	er Resets
'1' = Bit is se	t	'0' = Bit is clea	ared	HC = Bit is clear	ed by hardwar	e	
bit 7	Unimplemen	ted: Read as '1	,				
bit 6	CFGS: Config	guration Select b	bit				
		Configuration, U		ce ID Registers			
bit 5		Write Latches C	•				
				rite latch is loaded			
		ressed program vill be initiated o		atch is loaded/upd	ated and a wri	te of all program	n memory writ
bit 4				ommanu			
DIL 4		am Flash Erase s an erase opera		t WR command (I	nardware clea	red upon compl	etion)
		s a write operatio					,
bit 3		gram/Erase Erro					
		n indicates an in any set attempt		n or erase sequen	ce attempt or	termination (bit	is set automat
	-	gram or erase op					
bit 2		ram/Erase Enab	-	,			
	1 = Allows p	rogram/erase cy	cles				
	0 = Inhibits p	programming/era	ising of progran	n Flash			
bit 1	WR: Write Co						
		a program Flash		e operation.	are once one	ration is comple	te
		bit can only be					
				complete and ina	ctive		
bit 0	RD: Read Cor	ntrol bit					
			ead. Read takes	one cycle. RD is c	cleared in hard	ware. The RD bit	can only be se
	•	red) in software. initiate a prograr	n Flash read				
Note 1: U	Inimplemented bit,						
			by hardware whe	n a program memo	ory write or eras	se operation is st	arted (WR = 1
3 : T	he LWLO bit is ign	ored during a pro	gram memory e	rase operation (FR	EE = 1).		

REGISTER 10-5: PMCON1: PROGRAM MEMORY CONTROL 1 REGISTER

W-0/0	W-0/0	W-0/0	W-0/0	W-0/0	W-0/0	W-0/0	W-0/0			
Program Memory Control Register 2										
bit 7							bit 0			
Legend:										
R = Readable	bit	W = Writable	bit	U = Unimpler	mented bit, read	l as '0'				
S = Bit can onl	ly be set	x = Bit is unkr	nown	-n/n = Value a	at POR and BO	R/Value at all o	other Resets			
'1' = Bit is set		'0' = Bit is clea	ared							

REGISTER 10-6: PMCON2: PROGRAM MEMORY CONTROL 2 REGISTER

bit 7-0 Flash Memory Unlock Pattern bits

To unlock writes, a 55h must be written first, followed by an AAh, before setting the WR bit of the PMCON1 register. The value written to this register is used to unlock the writes. There are specific timing requirements on these writes.

TABLE 10-3: SUMMARY OF REGISTERS ASSOCIATED WITH FLASH PROGRAM MEMORY

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	132
PMCON1	_(1)	CFGS	LWLO	FREE	WRERR	WREN	WR	RD	171
PMCON2	Program Memory Control Register 2								172
PMADRL	PMADRL<7:0>								169
PMADRH	_(1)			Р	MADRH<6:()>			170
PMDATL	PMDATL<7:0>								169
PMDATH		— — PMDATH<5:0>							169

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by Flash program memory. **Note 1:** Unimplemented, read as '1'.

TABLE 10-4: SUMMARY OF CONFIGURATION WORD WITH FLASH PROGRAM MEMORY

Name	Bits	Bit -/7	Bit -/6	Bit 13/5	Bit 12/4	Bit 11/3	Bit 10/2	Bit 9/1	Bit 8/0	Register on Page	
CONFIG1	13:8	_		FCMEN	IESO	CLKOUTEN	BOREI	N<1:0>		95	
CONFIGI	7:0	CP	MCLRE	PWRTE	WDT	E<1:0>		FOSC<2:0>		90	
CONFIG2	13:8	_	—	LVP	DEBUG	LPBOR	BORV	STVREN	PLLEN	07	
	7:0	ZCD	_		_		PPS1WAY	WRT	<1:0>	97	

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by Flash program memory.

11.0 I/O PORTS

Each port has six standard registers for its operation. These registers are:

- TRISx registers (data direction)
- PORTx registers (reads the levels on the pins of the device)
- LATx registers (output latch)
- INLVLx (input level control)
- ODCONx registers (open-drain)
- · SLRCONx registers (slew rate

Some ports may have one or more of the following additional registers. These registers are:

- ANSELx (analog select)
- WPUx (weak pull-up)

In general, when a peripheral is enabled on a port pin, that pin cannot be used as a general purpose output. However, the pin can still be read.

TABLE 11-1: PORT AVAILABILITY PER DEVICE

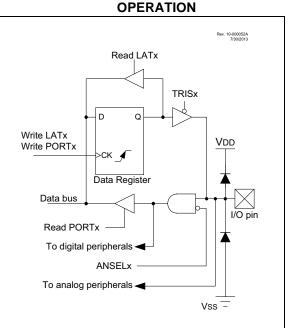
Device	PORTA	PORTB	PORTC	PORTD	PORTE
PIC16(L)F1778	•	•	•		٠
PIC16(L)F1777/9	•	•	•	٠	•

The Data Latch (LATx registers) is useful for read-modify-write operations on the value that the I/O pins are driving.

A write operation to the LATx register has the same effect as a write to the corresponding PORTx register. A read of the LATx register reads of the values held in the I/O PORT latches, while a read of the PORTx register reads the actual I/O pin value.

Ports that support analog inputs have an associated ANSELx register. When an ANSEL bit is set, the digital input buffer associated with that bit is disabled. Disabling the input buffer prevents analog signal levels on the pin between a logic high and low from causing excessive current in the logic input circuitry. A simplified model of a generic I/O port, without the interfaces to other peripherals, is shown in Figure 11-1.

FIGURE 11-1: GENERIC I/O PORT



11.1 PORTA Registers

11.1.1 DATA REGISTER

PORTA is a 6-bit wide, bidirectional port. The corresponding data direction register is TRISA (Register 11-2). Setting a TRISA bit (= 1) will make the corresponding PORTA pin an input (i.e., disable the output driver). Clearing a TRISA bit (= 0) will make the corresponding PORTA pin an output (i.e., enables output driver and puts the contents of the output latch on the selected pin). The exception is RA3, which is input-only and its TRIS bit will always read as '1'. Example 11-1 shows how to initialize PORTA.

Reading the PORTA register (Register 11-1) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch (LATA).

11.1.2 DIRECTION CONTROL

The TRISA register (Register 11-2) controls the PORTA pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISA register are maintained set when using them as analog inputs. I/O pins configured as analog inputs always read '0'.

11.1.3 OPEN-DRAIN CONTROL

The ODCONA register (Register 11-6) controls the open-drain feature of the port. Open-drain operation is independently selected for each pin. When an ODCONA bit is set, the corresponding port output becomes an open-drain driver capable of sinking current only. When an ODCONA bit is cleared, the corresponding port output pin is the standard push-pull drive capable of sourcing and sinking current.

11.1.4 SLEW RATE CONTROL

The SLRCONA register (Register 11-7) controls the slew rate option for each port pin. Slew rate control is independently selectable for each port pin. When an SLRCONA bit is set, the corresponding port pin drive is slew rate limited. When an SLRCONA bit is cleared, The corresponding port pin drive slews at the maximum rate possible.

11.1.5 INPUT THRESHOLD CONTROL

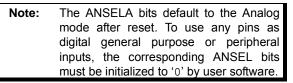
The INLVLA register (Register 11-8) controls the input voltage threshold for each of the available PORTA input pins. A selection between the Schmitt Trigger CMOS or the TTL Compatible thresholds is available. The input threshold is important in determining the value of a read of the PORTA register and also the level at which an interrupt-on-change occurs, if that feature is enabled. See Table 36-4: I/O Ports for more information on threshold levels.

Note: Changing the input threshold selection should be performed while all peripheral modules are disabled. Changing the threshold level during the time a module is active may inadvertently generate a transition associated with an input pin, regardless of the actual voltage level on that pin.

11.1.6 ANALOG CONTROL

The ANSELA register (Register 11-4) is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSELA bit high will cause all digital reads on the pin to be read as '0' and allow analog functions on the pin to operate correctly.

The state of the ANSELA bits has no effect on digital output functions. A pin with TRIS clear and ANSEL set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.



EXAMPLE 11-1: INITIALIZING PORTA

	ode example								
; initia	; initializing the PORTA register. The								
; other	; other ports are initialized in the same								
; manner									
BANKSEL	PORTA	;							
CLRF	PORTA	;Init PORTA							
BANKSEL	LATA	;Data Latch							
CLRF	LATA	;							
BANKSEL	ANSELA	;							
CLRF	ANSELA	;digital I/O							
BANKSEL	TRISA	;							
MOVLW	B'00111000'	;Set RA<5:3> as inputs							
MOVWF	TRISA	;and set RA<2:0> as							
		;outputs							

11.1.7 PORTA FUNCTIONS AND OUTPUT PRIORITIES

Each PORTA pin is multiplexed with other functions.

Each pin defaults to the PORT latch data after reset. Other functions are selected with the peripheral pin select logic. See **Section 12.0** "**Peripheral Pin Select (PPS) Module**" for more information.

Analog input functions, such as ADC and comparator inputs are not shown in the peripheral pin select lists. These inputs are active when the I/O pin is set for Analog mode using the ANSELA register. Digital output functions may continue to control the pin when it is in Analog mode.

11.2 Register Definitions: PORTA

REGISTER 11-1: PORTA: PORTA REGISTER

R/W-x/x	R/W-x/x	R/W-x/x	R/W-x/x	R-x/x	R/W-x/x	R/W-x/x	R/W-x/x	
RA7	RA6	RA5	RA4	RA3	RA2	RA1	RA0	
bit 7	·	·					bit 0	
Legend:								
R = Readable	bit	W = Writable	bit	U = Unimplemented bit, read as '0'				
u = Bit is unch	u = Bit is unchanged x = Bit is unknown			-n/n = Value at POR and BOR/Value at all other Resets				
'1' = Bit is set		'0' = Bit is clea	ared					

bit 7-0 RA<7:0>: PORTA I/O Value bits⁽¹⁾ 1 = Port pin is ≥ VIH 0 = Port pin is ≤ VIL

Note 1: Writes to PORTA are actually written to corresponding LATA register. Reads from PORTA register is return of actual I/O pin values.

REGISTER 11-2: TRISA: PORTA TRI-STATE REGISTER

| R/W-1/1 |
|---------|---------|---------|---------|---------|---------|---------|---------|
| TRISA7 | TRISA6 | TRISA5 | TRISA4 | TRISA3 | TRISA2 | TRISA1 | TRISA0 |
| bit 7 | | | • | • | | | bit 0 |

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0

TRISA<7:0>: PORTA Tri-State Control bit

1 = PORTA pin configured as an input (tri-stated)

0 = PORTA pin configured as an output

REGISTER 11-3: LATA: PORTA DATA LATCH REGISTER

| R/W-x/u |
|---------|---------|---------|---------|---------|---------|---------|---------|
| LATA7 | LATA6 | LATA5 | LATA4 | LATA3 | LATA2 | LATA1 | LATA0 |
| bit 7 | | | | | | | bit 0 |

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 LATA<7:0>: RA<7:0> Output Latch Value bits⁽¹⁾

Note 1: Writes to PORTA are actually written to corresponding LATA register. Reads from PORTA register is return of actual I/O pin values.

U-0	U-0	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1
_	—	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0
bit 7						•	bit 0
Legend:							
R = Readable bit W = Writable bit			U = Unimplemented bit, read as '0'				
u = Bit is uncha	= Bit is unchanged x = Bit is unknown		-n/n = Value at POR and BOR/Value at all other Resets				
'1' = Bit is set '0' = Bit is cleared							

REGISTER 11-4: ANSELA: PORTA ANALOG SELECT REGISTER

bit 7-6	Unimplemented: Read as '0'
bit 5-0	ANSA<5:0>: Analog Select between Analog or Digital Function on pins RA<5:0>
	1 = Analog input. Pin is assigned as analog input ⁽¹⁾ . Digital input buffer disabled.
	0 = Digital I/O. Pin is assigned to port or digital special function.

Note 1: When setting a pin to an analog input, the corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.

REGISTER 11-5: WPUA: WEAK PULL-UP PORTA REGISTER

| R/W-1/1 |
|---------|---------|---------|---------|---------|---------|---------|---------|
| WPUA7 | WPUA6 | WPUA5 | WPUA4 | WPUA3 | WPUA2 | WPUA1 | WPUA0 |
| bit 7 | | | | | | | bit 0 |

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 WPUA<7:0>: Weak Pull-up Register bits^{(1),(2)}

- 1 = Pull-up enabled
- 0 = Pull-up disabled
- Note 1: Global WPUEN bit of the OPTION_REG register must be cleared for individual pull-ups to be enabled.
 - **2:** The weak pull-up device is automatically disabled if the pin is configured as an output.

| R/W-0/0 |
|---------|---------|---------|---------|---------|---------|---------|---------|
| ODA7 | ODA6 | ODA5 | ODA4 | ODA3 | ODA2 | ODA1 | ODA0 |
| bit 7 | | | | | | | bit 0 |
| | | | | | | | |

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **ODA<7:0>:** PORTA Open-Drain Enable bits For RA<7:0> pins

1 = Port pin operates as open-drain drive (sink current only)

0 = Port pin operates as standard push-pull drive (source and sink current)

REGISTER 11-7: SLRCONA: PORTA SLEW RATE CONTROL REGISTER

| R/W-1/1 |
|---------|---------|---------|---------|---------|---------|---------|---------|
| SLRA7 | SLRA6 | SLRA5 | SLRA4 | SLRA3 | SLRA2 | SLRA1 | SLRA0 |
| bit 7 | • | | | • | | | bit 0 |

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 SLRA<7:0>: PORTA Slew Rate Enable bits For RA<7:0> pins 1 = Port pin slew rate is limited

0 = Port pin slews at maximum rate

REGISTER 11-8: INLVLA: PORTA INPUT LEVEL CONTROL REGISTER

| R/W-0/0 |
|---------|---------|---------|---------|---------|---------|---------|---------|
| INLVLA7 | INLVLA6 | INLVLA5 | INLVLA4 | INLVLA3 | INLVLA2 | INLVLA1 | INLVLA0 |
| bit 7 | | | | • | | | bit 0 |

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 INLVLA<7:0>: PORTA Input Level Select bits

For RA<7:0> pins

1 = Port pin digital input operates with ST thresholds

0 = Port pin digital input operates with TTL thresholds

					-		-	_	_
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ANSELA	—	—	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0	177
INLVLA	INLVLA7	INLVLA6	INLVLA5	INLVLA4	INLVLA3	INLVLA2	INLVLA1	INLVLA0	178
LATA	LATA7	LATA6	LATA5	LATA4	LATA3	LATA2	LATA1	LATA0	176
ODCONA	ODA7	ODA6	ODA5	ODA4	ODA3	ODA2	ODA1	ODA0	178
OPTION_REG	WPUEN	INTEDG	TMR0CS	TMR0SE	PSA	PS<2:0>		274	
PORTA	RA7	RA6	RA5	RA4	RA3	RA2	RA1	RA0	176
SLRCONA	SLRA7	SLRA6	SLRA5	SLRA4	SLRA3	SLRA2	SLRA1	SLRA0	178
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	176
WPUA	WPUA7	WPUA6	WPUA5	WPUA4	WPUA3	WPUA2	WPUA1	WPUA0	177

TABLE 11-2: SUMMARY OF REGISTERS ASSOCIATED WITH PORTA

Legend: x = unknown, u = unchanged, – = unimplemented locations read as '0'. Shaded cells are not used by PORTA.

Name	Bits	Bit -/7	Bit -/6	Bit 13/5	Bit 12/4	Bit 11/3	Bit 10/2	Bit 9/1	Bit 8/0	Register on Page
CONFIG1	13:8			FCMEN	IESO	CLKOUTEN	BOREN<1:0>			95
CONFIGI	7:0	CP	MCLRE	PWRTE	WDTE<1:0>		FOSC<2:0>			95

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by PORTA.

11.3 PORTB Registers

11.3.1 DATA REGISTER

PORTB is an 8-bit wide, bidirectional port. The corresponding data direction register is TRISB (Register 11-10). Setting a TRISB bit (= 1) will make the corresponding PORTB pin an input (i.e., put the corresponding output driver in a High-Impedance mode). Clearing a TRISB bit (= 0) will make the corresponding PORTB pin an output (i.e., enable the output driver and put the contents of the output latch on the selected pin). Example 11-1 shows how to initialize an I/O port.

Reading the PORTB register (Register 11-9) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch (LATB).

11.3.2 DIRECTION CONTROL

The TRISB register (Register 11-10) controls the PORTB pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISB register are maintained set when using them as analog inputs. I/O pins configured as analog inputs always read '0'.

11.3.3 OPEN-DRAIN CONTROL

The ODCONB register (Register 11-14) controls the open-drain feature of the port. Open-drain operation is independently selected for each pin. When an ODCONB bit is set, the corresponding port output becomes an open-drain driver capable of sinking current only. When an ODCONB bit is cleared, the corresponding port output pin is the standard push-pull drive capable of sourcing and sinking current.

11.3.4 SLEW RATE CONTROL

The SLRCONB register (Register 11-15) controls the slew rate option for each port pin. Slew rate control is independently selectable for each port pin. When an SLRCONB bit is set, the corresponding port pin drive is slew rate limited. When an SLRCONB bit is cleared, The corresponding port pin drive slews at the maximum rate possible.

11.3.5 INPUT THRESHOLD CONTROL

The INLVLB register (Register 11-16) controls the input voltage threshold for each of the available PORTB input pins. A selection between the Schmitt Trigger CMOS or the TTL Compatible thresholds is available. The input threshold is important in determining the value of a read of the PORTB register and also the level at which an interrupt-on-change occurs, if that feature is enabled. See Table 36-4: I/O Ports for more information on threshold levels.

Note: Changing the input threshold selection should be performed while all peripheral modules are disabled. Changing the threshold level during the time a module is active may inadvertently generate a transition associated with an input pin, regardless of the actual voltage level on that pin.

11.3.6 ANALOG CONTROL

The ANSELB register (Register 11-12) is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSELB bit high will cause all digital reads on the pin to be read as '0' and allow analog functions on the pin to operate correctly.

The state of the ANSELB bits has no effect on digital output functions. A pin with TRIS clear and ANSELB set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

Note:	The ANSELB bits default to the Analog mode after reset. To use any pins as
	digital general purpose or peripheral
	inputs, the corresponding ANSEL bits
	must be initialized to '0' by user software.

11.3.7 PORTB FUNCTIONS AND OUTPUT PRIORITIES

Each pin defaults to the PORT latch data after reset. Other functions are selected with the peripheral pin select logic. See **Section 12.0** "**Peripheral Pin Select (PPS) Module**" for more information. Analog input functions, such as ADC and op amp inputs, are not shown in the peripheral pin select lists. These inputs are active when the I/O pin is set for Analog mode using the ANSELB register. Digital output functions may continue to control the pin when it is in Analog mode.

11.3.8 HIGH CURRENT DRIVE CONTROL

The output drivers on RB1 and RB0 are capable of sourcing and sinking up to 100 mA. This extra drive capacity can be enabled and disabled with the control bits in the HIDRVB register (Register 11-17).

11.4 Register Definitions: PORTB

REGISTER 11-9: PORTB: PORTB REGISTER

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	
RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	
bit 7 bit C							bit 0	
Legend:								
R = Readable	bit	W = Writable	bit	U = Unimplemented bit, read as '0'				
u = Bit is unch	anged	x = Bit is unkr	nown	-n/n = Value at POR and BOR/Value at all other Resets				
(1' = Bit is set (0' = Bit is cleared								

bit 7-0 **RB<7:0>**: PORTB General Purpose I/O Pin bits⁽¹⁾ 1 = Port pin is ≥ VIH 0 = Port pin is ≤ VIL

REGISTER 11-10: TRISB: PORTB TRI-STATE REGISTER

R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1
TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0
bit 7	bit 7 bit 0						bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0

- TRISB<7:0>: PORTB Tri-State Control bits
- 1 = PORTB pin configured as an input (tri-stated)
- 0 = PORTB pin configured as an output

REGISTER 11-11: LATB: PORTB DATA LATCH REGISTER

| R/W-x/u |
|---------|---------|---------|---------|---------|---------|---------|---------|
| LATB7 | LATB6 | LATB5 | LATB4 | LATB3 | LATB2 | LATB1 | LATB0 |
| bit 7 | | | | | | | bit 0 |

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 LATB<7:0>: PORTB Output Latch Value bits⁽¹⁾

Note 1: Writes to PORTB are actually written to corresponding LATB register. Reads from PORTB register is return of actual I/O pin values.

Note 1: Writes to PORTB are actually written to corresponding LATB register. Reads from PORTB register is return of actual I/O pin values.

U-0	U-0	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1
—	—	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0
bit 7							bit C
Legend:							
R = Readable	bit	W = Writable bit		U = Unimplemented bit, read as '0'			
u = Bit is unch	anged	x = Bit is unkr	nown	-n/n = Value a	at POR and BO	R/Value at all o	other Resets
'1' = Bit is set		'0' = Bit is clea	ared				
bit 7-6	Unimplemen	ted: Read as '	0'				

bit 5-0 **ANSB<5:0>**: Analog Select between Analog or Digital Function on pins RB<5:0> 1 = Analog input. Pin is assigned as analog input⁽¹⁾. Digital input buffer disabled. 0 = Digital I/O. Pin is assigned to port or digital special function.

Note 1: When setting a pin to an analog input, the corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.

REGISTER 11-13: WPUB: WEAK PULL-UP PORTB REGISTER

| R/W-1/1 |
|---------|---------|---------|---------|---------|---------|---------|---------|
| WPUB7 | WPUB6 | WPUB5 | WPUB4 | WPUB3 | WPUB2 | WPUB1 | WPUB0 |
| bit 7 | | | | | | | bit 0 |

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-4 **WPUB<7:0>**: Weak Pull-up Register bits^(1,2)

1 = Pull-up enabled

0 = Pull-up disabled

- Note 1: Global WPUEN bit of the OPTION_REG register must be cleared for individual pull-ups to be enabled.
 - 2: The weak pull-up device is automatically disabled if the pin is configured as an output.

bit 7 bit 0						

REGISTER 11-14: ODCONB: PORTB OPEN-DRAIN CONTROL REGISTER

bit 7-0 **ODB<7:0>:** PORTB Open-Drain Enable bits For RB<7:0> pins

'1' = Bit is set

1 = Port pin operates as open-drain drive (sink current only)

0 = Port pin operates as standard push-pull drive (source and sink current)

REGISTER 11-15: SLRCONB: PORTB SLEW RATE CONTROL REGISTER

'0' = Bit is cleared

R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1
SLRB7	SLRB6	SLRB5	SLRB4	SLRB3	SLRB2	SLRB1	SLRB0
bit 7	bit 7 bit 0						

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 SLRB<7:0>: PORTB Slew Rate Enable bits

For RB<7:0> pins

1 = Port pin slew rate is limited

0 = Port pin slews at maximum rate

REGISTER 11-16: INLVLB: PORTB INPUT LEVEL CONTROL REGISTER

| R/W-0/0 |
|---------|---------|---------|---------|---------|---------|---------|---------|
| INLVLB7 | INLVLB6 | INLVLB5 | INLVLB4 | INLVLB3 | INLVLB2 | INLVLB1 | INLVLB0 |
| bit 7 | | | | | | | bit 0 |

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0

INLVLB<7:0>: PORTB Input Level Select bits

For RB<7:0> pins

1 = Port pin digital input operates with ST thresholds

0 = Port pin digital input operates with TTL thresholds

REGISTER 11-17: HIDRVB: PORTB HIGH DRIVE CONTROL REGISTER

U-0	U-0	U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0
—			—	—	—	HIDB1	HIDB0
bit 7 bit 0							

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-2 Unimplemented: Read as '0'

bit 1-0 HIDB<1:0>: PORTB High Drive Enable bits For RB<1:0> pins 1 = High current source and sink enabled

0 = Standard current source and sink

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ANSELB	—	—	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	182
HIDRVB	—	—	-	—	—	—	HIDB1	HIDB0	184
INLVLB	INLVLB7	INLVLB6	INLVLB5	INLVLB4	INLVLB3	INLVLB2	INLVLB1	INLVLB0	183
LATB	LATB7	LATB6	LATB5	LATB4	LATB3	LATB2	LATB1	LATB0	181
ODCONB	ODB7	ODB6	ODB5	ODB4	ODB3	ODB2	ODB1	ODB0	183
PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	181
SLRCONB	SLRB7	SLRB6	SLRB5	SLRB4	SLRB3	SLRB2	SLRB1	SLRB0	183
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	183
WPUB	WPUB7	WPUB6	WPUB5	WPUB4	WPUB3	WPUB2	WPUB1	WPUB0	182

TABLE 11-4: SUMMARY OF REGISTERS ASSOCIATED WITH PORTB

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by PORTB.

11.5 PORTC Registers

11.5.1 DATA REGISTER

PORTC is an 8-bit wide bidirectional port in the PIC16(L)F1777/8/9 devices. The corresponding data direction register is TRISC (Register 11-19). Setting a TRISC bit (= 1) will make the corresponding PORTC pin an input (i.e., put the corresponding output driver in a High-Impedance mode). Clearing a TRISC bit (= 0) will make the corresponding PORTC pin an output (i.e., enable the output driver and put the contents of the output latch on the selected pin). Example 11-1 shows how to initialize an I/O port.

Reading the PORTC register (Register 11-18) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch (LATC).

11.5.2 DIRECTION CONTROL

The TRISC register (Register 11-19) controls the PORTC pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISC register are maintained set when using them as analog inputs. I/O pins configured as analog inputs always read '0'.

11.5.3 INPUT THRESHOLD CONTROL

The INLVLC register (Register 11-25) controls the input voltage threshold for each of the available PORTC input pins. A selection between the Schmitt Trigger CMOS or the TTL Compatible thresholds is available. The input threshold is important in determining the value of a read of the PORTC register and also the level at which an interrupt-on-change occurs, if that feature is enabled. See Table 36-4: I/O Ports for more information on threshold levels.

Note: Changing the input threshold selection should be performed while all peripheral modules are disabled. Changing the threshold level during the time a module is active may inadvertently generate a transition associated with an input pin, regardless of the actual voltage level on that pin.

11.5.4 OPEN-DRAIN CONTROL

The ODCONC register (Register 11-23) controls the open-drain feature of the port. Open-drain operation is independently selected for each pin. When an ODCONC bit is set, the corresponding port output becomes an open-drain driver capable of sinking current only. When an ODCONC bit is cleared, the corresponding port output pin is the standard push-pull drive capable of sourcing and sinking current.

11.5.5 SLEW RATE CONTROL

The SLRCONC register (Register 11-24) controls the slew rate option for each port pin. Slew rate control is independently selectable for each port pin. When an SLRCONC bit is set, the corresponding port pin drive is slew rate limited. When an SLRCONC bit is cleared, The corresponding port pin drive slews at the maximum rate possible.

11.5.6 ANALOG CONTROL

The ANSELC register (Register 11-21) is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSELC bit high will cause all digital reads on the pin to be read as '0' and allow analog functions on the pin to operate correctly.

The state of the ANSELC bits has no effect on digital output functions. A pin with TRIS clear and ANSELC set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

Note:	The ANSELC bits default to the Analog			
	mode after reset. To use any pins as			
	digital general purpose or peripheral			
	inputs, the corresponding ANSEL bits			
	must be initialized to '0' by user software.			

11.5.7 PORTC FUNCTIONS AND OUTPUT PRIORITIES

Each pin defaults to the PORT latch data after reset. Other functions are selected with the peripheral pin select logic. See **Section 12.0** "**Peripheral Pin Select (PPS) Module**" for more information.

Analog input functions, such as ADC and comparator inputs, are not shown in the peripheral pin select lists. These inputs are active when the I/O pin is set for Analog mode using the ANSELC register. Digital output functions may continue to control the pin when it is in Analog mode.

11.6 Register Definitions: PORTC

REGISTER 11-18: PORTC: PORTC REGISTER

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	
RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0	
bit 7							bit 0	
Legend:								
R = Readable	R = Readable bit W = Writable bit			U = Unimplemented bit, read as '0'				
u = Bit is unchanged x = Bit is unknown		-n/n = Value at POR and BOR/Value at all other Resets						
'1' = Bit is set '0' = Bit is cleared								

bit 7-0 **RC<7:0>**: PORTC General Purpose I/O Pin bits⁽¹⁾ 1 = Port pin is ≥ VIH 0 = Port pin is ≤ VIL

Note 1: Writes to PORTC are actually written to corresponding LATC register. Reads from PORTC register is return of actual I/O pin values.

REGISTER 11-19: TRISC: PORTC TRI-STATE REGISTER

| R/W-1/1 |
|---------|---------|---------|---------|---------|---------|---------|---------|
| TRISC7 | TRISC6 | TRISC5 | TRISC4 | TRISC3 | TRISC2 | TRISC1 | TRISC0 |
| bit 7 | | | | | | | bit 0 |

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **TRISC<7:0>:** PORTC Tri-State Control bits 1 = PORTC pin configured as an input (tri-stated) 0 = PORTC pin configured as an output

REGISTER 11-20:	LATC: PORTC DATA LATCH REGISTER
------------------------	---------------------------------

| R/W-x/u |
|---------|---------|---------|---------|---------|---------|---------|---------|
| LATC7 | LATC6 | LATC5 | LATC4 | LATC3 | LATC2 | LATC1 | LATC0 |
| bit 7 | | | | | | | bit 0 |

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 LATC<7:0>: PORTC Output Latch Value bits

REGISTER 11-21: ANSELC: PORTC ANALOG SELECT REGISTER

R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	U-0	U-0
ANSC7	ANSC6	ANSC5	ANSC4	ANSC3	ANSC2	—	—
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-2 ANSC<7:2>: Analog Select between Analog or Digital Function on pins RC<7:2>⁽¹⁾

1 = Analog input. Pin is assigned as analog input⁽¹⁾. Digital input buffer disabled.

0 = Digital I/O. Pin is assigned to port or digital special function.

bit 1-0 Unimplemented: Read as '0'

Note 1: When setting a pin to an analog input, the corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.

REGISTER 11-22: WPUC: WEAK PULL-UP PORTC REGISTER

Legend:							
bit 7							bit 0
WPUC7	WPUC6	WPUC5	WPUC4	WPUC3	WPUC2	WPUC1	WPUC0
R/W-1/1							

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 WPUC<7:0>: Weak Pull-up Register bits^(1, 2) 1 = Pull-up enabled 0 = Pull-up disabled

Note 1: Global WPUEN bit of the OPTION_REG register must be cleared for individual pull-ups to be enabled.

2: The weak pull-up device is automatically disabled if the pin is configured as an output.

REGISTER 11-23: ODCONC: PORTC OPEN-DRAIN CONTROL REGISTER

| R/W-0/0 |
|---------|---------|---------|---------|---------|---------|---------|---------|
| ODC7 | ODC6 | ODC5 | ODC4 | ODC3 | ODC2 | ODC1 | ODC0 |
| bit 7 | | • | | | | | bit 0 |
| | | | | | | | |
| Legend: | | | | | | | |

Legenu.		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **ODC<7:0>:** PORTC Open-Drain Enable bits

For RC<7:0> pins

- 1 = Port pin operates as open-drain drive (sink current only)
- 0 = Port pin operates as standard push-pull drive (source and sink current)

Legend:							
bit 7							bit 0
SLRC7	SLRC6	SLRC5	SLRC4	SLRC3	SLRC2	SLRC1	SLRC0
R/W-1/1							

REGISTER 11-24: SLRCONC: PORTC SLEW RATE CONTROL REGISTER

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 SLRC<7:0>: PORTC Slew Rate Enable bits For RC<7:0> pins 1 = Port pin slew rate is limited 0 = Port pin slews at maximum rate

REGISTER 11-25: INLVLC: PORTC INPUT LEVEL CONTROL REGISTER

| R/W-1/1 |
|---------|---------|---------|---------|---------|---------|---------|---------|
| INLVLC7 | INLVLC6 | INLVLC5 | INLVLC4 | INLVLC3 | INLVLC2 | INLVLC1 | INLVLC0 |
| bit 7 | | | | | | | bit 0 |

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0

INLVLC<7:0>: PORTC Input Level Select bits

For RC<7:0> pins

1 = Port pin digital input operates with ST thresholds

0 = Port pin digital input operates with TTL thresholds

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Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ANSELC	ANSC7	ANSC6	ANSC5	ANSC4	ANSC3	ANSC2	_	_	187
INLVLC	INLVLC7	INLVLC6	INLVLC5	INLVLC4	INLVLC3	INLVLC2	INLVLC1	INLVLC0	189
LATC	LATC7	LATC6	LATC5	LATC4	LATC3	LATC2	LATC1	LATC0	187
ODCONC	ODC7	ODC6	ODC5	ODC4	ODC3	ODC2	ODC1	ODC0	188
PORTC	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0	186
SLRCONC	SLRC7	SLRC6	SLRC5	SLRC4	SLRC3	SLRC2	SLRC1	SLRC0	189
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	186
WPUC	WPUC7	WPUC6	WPUC5	WPUC4	WPUC3	WPUC2	WPUC1	WPUC0	188

TABLE 11-5: SUMMARY OF REGISTERS ASSOCIATED WITH PORTC

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by PORTC.

11.7 PORTD Registers (PIC16(L)F1777/9 only)

11.7.1 DATA REGISTER

PORTD is an 8-bit wide bidirectional port in the PIC16(L)F1777/8/9 devices. The corresponding data direction register is TRISD (Register 11-27). Setting a TRISD bit (= 1) will make the corresponding PORTD pin an input (i.e., put the corresponding output driver in a High-Impedance mode). Clearing a TRISD bit (= 0) will make the corresponding PORTD pin an output (i.e., enable the output driver and put the contents of the output latch on the selected pin). Example 11-1 shows how to initialize an I/O port.

Reading the PORTD register (Register 11-26) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch (LATD).

11.7.2 DIRECTION CONTROL

The TRISD register (Register 11-27) controls the PORTD pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISD register are maintained set when using them as analog inputs. I/O pins configured as analog inputs always read '0'.

11.7.3 INPUT THRESHOLD CONTROL

The INLVLD register (Register 11-33) controls the input voltage threshold for each of the available PORTD input pins. A selection between the Schmitt Trigger CMOS or the TTL Compatible thresholds is available. The input threshold is important in determining the value of a read of the PORTD register and also the level at which an interrupt-on-change occurs, if that feature is enabled. See Table 36-4: I/O Ports for more information on threshold levels.

Note: Changing the input threshold selection should be performed while all peripheral modules are disabled. Changing the threshold level during the time a module is active may inadvertently generate a transition associated with an input pin, regardless of the actual voltage level on that pin.

11.7.4 OPEN-DRAIN CONTROL

The ODCOND register (Register 11-31) controls the open-drain feature of the port. Open-drain operation is independently selected for each pin. When an ODCOND bit is set, the corresponding port output becomes an open-drain driver capable of sinking current only. When an ODCOND bit is cleared, the corresponding port output pin is the standard push-pull drive capable of sourcing and sinking current.

11.7.5 SLEW RATE CONTROL

The SLRCOND register (Register 11-32) controls the slew rate option for each port pin. Slew rate control is independently selectable for each port pin. When an SLRCOND bit is set, the corresponding port pin drive is slew rate limited. When an SLRCOND bit is cleared, The corresponding port pin drive slews at the maximum rate possible.

11.7.6 ANALOG CONTROL

The ANSELD register (Register 11-29) is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSELD bit high will cause all digital reads on the pin to be read as '0' and allow analog functions on the pin to operate correctly.

The state of the ANSELD bits has no effect on digital output functions. A pin with TRIS clear and ANSELD set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

Note:	The ANSELD bits default to the Analog
	mode after reset. To use any pins as
	digital general purpose or peripheral
	inputs, the corresponding ANSEL bits
	must be initialized to '0' by user software.

11.7.7 PORTD FUNCTIONS AND OUTPUT PRIORITIES

Each pin defaults to the PORT latch data after reset. Other functions are selected with the peripheral pin select logic. See **Section 12.0** "**Peripheral Pin Select (PPS) Module**" for more information.

Analog input functions, such as ADC and comparator inputs, are not shown in the peripheral pin select lists. These inputs are active when the I/O pin is set for Analog mode using the ANSELD register. Digital output functions may continue to control the pin when it is in Analog mode.

11.8 Register Definitions: PORTD

REGISTER 11-26: PORTD: PORTD REGISTER

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0
bit 7							bit 0
Legend:							
R = Readable	R = Readable bit W = Writable bit		U = Unimplemented bit, read as '0'				
u = Bit is uncha	anged	ed x = Bit is unknown		-n/n = Value at POR and BOR/Value at all other Resets			
'1' = Bit is set '0' = Bit is cleared							

bit 7-0 **RD<7:0>**: PORTD General Purpose I/O Pin bits⁽¹⁾ 1 = Port pin is ≥ VIH 0 = Port pin is ≤ VIL

Note 1: Writes to PORTD are actually written to corresponding LATD register. Reads from PORTD register is return of actual I/O pin values.

REGISTER 11-27: TRISD: PORTD TRI-STATE REGISTER

| R/W-1/1 |
|---------|---------|---------|---------|---------|---------|---------|---------|
| TRISD7 | TRISD6 | TRISD5 | TRISD4 | TRISD3 | TRISD2 | TRISD1 | TRISD0 |
| bit 7 | | | | | | | bit 0 |

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **TRISD<7:0>:** PORTD Tri-State Control bits 1 = PORTD pin configured as an input (tri-stated) 0 = PORTD pin configured as an output

LATD7 LATD6 LATD5 LATD4 LATD3 LATD2 LATE	
	D1 LATD0
bit 7	bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 LATD<7:0>: PORTD Output Latch Value bits

REGISTER 11-29: ANSELD: PORTD ANALOG SELECT REGISTER

| R/W-1/1 |
|---------|---------|---------|---------|---------|---------|---------|---------|
| ANSD7 | ANSD6 | ANSD5 | ANSD4 | ANSD3 | ANSD2 | ANSD1 | ANSD0 |
| bit 7 | | | | | | | bit 0 |

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **ANSD<7:0>**: Analog Select between Analog or Digital Function on pins RD<7:0>⁽¹⁾

1 = Analog input. Pin is assigned as analog input⁽¹⁾. Digital input buffer disabled.

0 = Digital I/O. Pin is assigned to port or digital special function.

Note 1: When setting a pin to an analog input, the corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.

REGISTER 11-30: WPUD: WEAK PULL-UP PORTD REGISTER

Legend:							
bit 7							bit 0
WPUD7	WPUD6	WPUD5	WPUD4	WPUD3	WPUD2	WPUD1	WPUD0
R/W-1/1							

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 WPUD<7:0>: Weak Pull-up Register bits^(1, 2) 1 = Pull-up enabled 0 = Pull-up disabled

Note 1: Global WPUEN bit of the OPTION_REG register must be cleared for individual pull-ups to be enabled.

2: The weak pull-up device is automatically disabled if the pin is configured as an output.

REGISTER 11-31: ODCOND: PORTD OPEN-DRAIN CONTROL REGISTER

| R/W-0/0 |
|---------|---------|---------|---------|---------|---------|---------|---------|
| ODD7 | ODD6 | ODD5 | ODD4 | ODD3 | ODD2 | ODD1 | ODD0 |
| bit 7 | | | | | | | bit 0 |
| | | | | | | | |
| Legend: | | | | | | | |

Ecgenia.		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **ODD<7:0>:** PORTD Open-Drain Enable bits

For RD<7:0> pins

- 1 = Port pin operates as open-drain drive (sink current only)
- 0 = Port pin operates as standard push-pull drive (source and sink current)

-n/n = Value at POR and BOR/Value at all other Resets

Legend:							
bit 7							bit 0
SLRD7	SLRD6	SLRD5	SLRD4	SLRD3	SLRD2	SLRD1	SLRD0
R/W-1/1							

REGISTER 11-32: SLRCOND: PORTD SLEW RATE CONTROL REGISTER

bit 7-0 SLRD<7:0>: PORTD Slew Rate Enable bits For RD<7:0> pins 1 = Port pin slew rate is limited 0 = Port pin slews at maximum rate

u = Bit is unchanged

'1' = Bit is set

REGISTER 11-33: INLVLD: PORTD INPUT LEVEL CONTROL REGISTER

x = Bit is unknown

'0' = Bit is cleared

| R/W-1/1 |
|---------|---------|---------|---------|---------|---------|---------|---------|
| INLVLD7 | INLVLD6 | INLVLD5 | INLVLD4 | INLVLD3 | INLVLD2 | INLVLD1 | INLVLD0 |
| bit 7 | | | | | | | bit 0 |

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 INLVLD<7:0>: PORTD Input Level Select bits

For RD<7:0> pins

1 = Port pin digital input operates with ST thresholds

0 = Port pin digital input operates with TTL thresholds

PIC16(L)F1777/8/9

TABLE 11-6:	SUMMARY OF REGISTERS ASSOCIATED WITH PORTD ⁽¹⁾
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Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ANSELD	ANSD7	ANSD6	ANSD5	ANSD4	ANSD3	ANSD2	ANSD1	ANSD0	193
INLVLD	INLVLD7	INLVLD6	INLVLD5	INLVLD4	INLVLD3	INLVLD2	INLVLD1	INLVLD0	195
LATD	LATD7	LATD6	LATD5	LATD4	LATD3	LATD2	LATD1	LATD0	193
ODCOND	ODD7	ODD6	ODD5	ODD4	ODD3	ODD2	ODD1	ODD0	194
PORTD	RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0	195
SLRCOND	SLRD7	SLRD6	SLRD5	SLRD4	SLRD3	SLRD2	SLRD1	SLRD0	195
TRISD	TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0	192
WPUD	WPUD7	WPUD6	WPUD5	WPUD4	WPUD3	WPUD2	WPUD1	WPUD0	194

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by PORTD.

Note 1: PIC16(L)F1777/9 only.

11.9 PORTE Registers

11.9.1 DATA REGISTER

PORTE is a 4-bit wide, bidirectional port. The corresponding data direction register is TRISE. Setting a TRISE bit (= 1) will make the corresponding PORTE pin an input (i.e., put the corresponding output driver in a High-Impedance mode). Clearing a TRISE bit (= 0) will make the corresponding PORTE pin an output (i.e., enable the output driver and put the contents of the output latch on the selected pin). The exception is RE3, which is input only and its TRIS bit will always read as '1'. Example 11-1 shows how to initialize an I/O port.

Reading the PORTE register (Register 11-34) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch (LATE). RE3 reads '0' when MCLRE = 1.

Note:	The RE<2:0>, TRISE<2:0>, INLVL<2:0>							
	and WPUE<2:0> pins are available on							
	PIC16(L)F1777/9 only.							

11.9.2 DIRECTION CONTROL

The TRISE register (Register 11-35) controls the PORTE pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISE register are maintained set when using them as analog inputs. I/O pins configured as analog inputs always read '0'.

11.9.3 INPUT THRESHOLD CONTROL

The INLVLE register (Register 11-41) controls the input voltage threshold for each of the available PORTE input pins. A selection between the Schmitt Trigger CMOS or the TTL Compatible thresholds is available. The input threshold is important in determining the value of a read of the PORTE register and also the level at which an interrupt-on-change occurs, if that feature is enabled. See Table 36-4: I/O Ports for more information on threshold levels.

Note: Changing the input threshold selection should be performed while all peripheral modules are disabled. Changing the threshold level during the time a module is active may inadvertently generate a transition associated with an input pin, regardless of the actual voltage level on that pin.

11.9.4 OPEN-DRAIN CONTROL⁽¹⁾

The ODCONE register (Register 11-39) controls the open-drain feature of the port. Open-drain operation is independently selected for each pin. When an ODCONE bit is set, the corresponding port output becomes an open-drain driver capable of sinking current only. When an ODCONE bit is cleared, the corresponding port output pin is the standard push-pull drive capable of sourcing and sinking current.

Note 1: Implemented on PIC16(L)F1777/9 only.

11.9.5 SLEW RATE CONTROL⁽¹⁾

The SLRCONE register (Register 11-40) controls the slew rate option for each port pin. Slew rate control is independently selectable for each port pin. When an SLRCONE bit is set, the corresponding port pin drive is slew rate limited. When an SLRCONE bit is cleared, The corresponding port pin drive slews at the maximum rate possible.

Note 1: Implemented on PIC16(L)F1777/9 only.

11.9.6 ANALOG CONTROL⁽¹⁾

The ANSELE register (Register 11-37) is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSELE bit high will cause all digital reads on the pin to be read as '0' and allow analog functions on the pin to operate correctly.

The state of the ANSELE bits has no effect on digital output functions. A pin with TRIS clear and ANSEL set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

The TRISE register (Register 11-35) controls the PORTE pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISE register are maintained set when using them as analog inputs. I/O pins configured as analog input always read '0'.

Note 1: Implemented on PIC16(L)F1777/9 only.

2: The ANSELE bits default to the Analog mode after Reset. To use any pins as digital general purpose or peripheral inputs, the corresponding ANSEL bits must be initialized to '0' by user software.

11.9.7 PORTE FUNCTIONS AND OUTPUT PRIORITIES

Each pin defaults to the PORT latch data after Reset. Other functions are selected with the peripheral pin select logic. See **Section 12.0** "**Peripheral Pin Select (PPS) Module**" for more information.

Analog input functions, such as ADC and comparator inputs, are not shown in the peripheral pin select lists. These inputs are active when the I/O pin is set for Analog mode using the ANSELE register. Digital output functions may continue to control the pin when it is in Analog mode.

11.10 Register Definitions: PORTE

REGISTER 11-34: PORTE: PORTE REGISTER

U-0	U-0	U-0	U-0	R-x/u	R/W-x/u	R/W-x/u	R/W-x/u		
—	_	_	—	RE3	RE2 ⁽¹⁾	RE1 ⁽¹⁾	RE0 ⁽¹⁾		
bit 7			•				bit 0		
Legend:									
R = Readable	e bit	W = Writable	bit	U = Unimplemented bit, read as '0'					
u = Bit is uncl	nanged	x = Bit is unkr	nown	-n/n = Value at POR and BOR/Value at all other Resets					
'1' = Bit is set '0' = Bit is cleared									
bit 7-4	Unimplemen	ted: Read as '	o'						
			(1)						

bit 3-0	RE<3:0>: PORTE I/O Pin bits ⁽¹⁾
	1 = Port pin is > Vін
	0 = Port pin is < VIL

Note 1: RE<2:0> are not implemented on the PIC16(L)F1778. Read as '0'. Writes to RE<2:0> are actually written to corresponding LATE register. Reads from PORTE register is the return of actual I/O pin values.

REGISTER 11-35: TRISE: PORTE TRI-STATE REGISTER

U-0	U-0	U-0	U-0	U-1 ⁽²⁾	R/W-1	R/W-1	R/W-1
—	—	_	—	—	TRISE2 ⁽¹⁾	TRISE1 ⁽¹⁾	TRISE0 ⁽¹⁾
bit 7				•			bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-4	Unimplemented: Read as '0'
bit 3	Unimplemented: Read as '1'
bit 2-0	TRISE<2:0>: RE<2:0> Tri-State Control bits ⁽¹⁾
	1 = PORTE pin configured as an input (tri-stated)
	0 = PORTE pin configured as an output

- **Note 1:** TRISE<2:0> are not implemented on the PIC16(L)F1778.
 - 2: Unimplemented, read as '1'.

REGISTER 11-36: LATE: PORTE DATA LATCH REGISTER⁽¹⁾

U-0	U-0	U-0	U-0	U-0	R/W-x/u	R/W-x/u	R/W-x/u
_	_	_	_	—	LATE2	LATE1	LATE0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-3 Unimplemented: Read as '0'

bit 2-0 LATE<2:0>: PORTE Output Latch Value bits

The LATE register is not implemented on the PIC16(L)F1778. Writes to RE<2:0> are actually written to Note 1: corresponding LATE register. Reads from PORTE register is the return of actual I/O pin values.

REGISTER 11-37: ANSELE: PORTE ANALOG SELECT REGISTER⁽²⁾

U-0	U-0	U-0	U-0	U-0	R/W-1/1	R/W-1/1	R/W-1/1
—	_	_	_	_	ANSE2	ANSE1	ANSE0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-3 Unimplemented: Read as '0'

bit 2-0

- **ANSE<2:0>**: Analog Select between Analog or Digital Function on pins RE<2:0>⁽¹⁾ 1 = Analog input. Pin is assigned as analog input⁽¹⁾. Digital input buffer disabled.

 - 0 = Digital I/O. Pin is assigned to port or digital special function.
- Note 1: When setting a pin to an analog input, the corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.
 - 2: The ANSELE register is not implemented on the PIC16(L)F1778.

REGISTER 11-38: WPUE: WEAK PULL-UP PORTE REGISTER

	U-0	U-0	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1
—		_	WPUE3	WPUE2 ⁽³⁾	WPUE1 ⁽³⁾	WPUE0 ⁽³⁾
						bit 0
	_			— — — WPUE3	— — — WPUE3 WPUE2 ⁽³⁾	— — — WPUE3 WPUE2 ⁽³⁾ WPUE1 ⁽³⁾

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-4	Unimplemented: Read as '0'
bit 3-0	WPUE<3:0>: Weak Pull-up Register bit ⁽³⁾
	1 = Pull-up enabled
	0 = Pull-up disabled

Note 1: Global WPUEN bit of the OPTION_REG register must be cleared for individual pull-ups to be enabled.

- **2:** The weak pull-up device is automatically disabled if the pin is in configured as an output.
- **3:** WPUE<2:0> is not implemented on the PIC16(L)F1778.

REGISTER 11-39: ODCONE: PORTE OPEN-DRAIN CONTROL REGISTER⁽¹⁾

U-0	U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0
_	—	—	_	_	ODE2	ODE1	ODE0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 2-0 **ODE<2:0>:** PORTE Open-Drain Enable bits For RE<2:0> pins 1 = Port pin operates as open-drain drive (sink current only) 0 = Port pin operates as standard push-pull drive (source and sink current)

Note 1: The ODCONE register is not implemented on the PIC16(L)F1778.

REGISTER 11-40: SLRCONE: PORTE SLEW RATE CONTROL REGISTER⁽¹⁾

U-0	U-0	U-0	U-0	U-0	R/W-1/1	R/W-1/1	R/W-1/1
—	—	—	—	_	SLRE2	SLRE1	SLRE0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-3 Unimplemented: Read as '0'

bit 2-0	SLRE<2:0>: PORTE Slew Rate Enable bits
	For RE<2:0> pins
	1 = Port pin slew rate is limited
	0 = Port pin slews at maximum rate

Note 1: The SLRCONE register is not implemented on the PIC16(L)F1778.

REGISTER 11-41: INLVLE: PORTE INPUT LEVEL CONTROL REGISTER

U-0	U-0	U-0	U-0	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1
_	—	_	—	INLVLE3	INLVLE2 ⁽¹⁾	INLVLE1 ⁽¹⁾	INLVLE0 ⁽¹⁾
bit 7	•					•	bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-4 Unimplemented: Read as '0'

bit 3-0 INLVLE<3:0>: PORTE Input Level Select bit 1 = ST input used for PORT reads and interrupt-on-change

0 = TTL input used for PORT reads and interrupt-on-change

Note 1: INLVLE<2:0> are not implemented on the PIC16(L)F1778.

TABLE 11-7: SUMMARY OF REGISTERS ASSOCIATED WITH PORTE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ANSELE ⁽²⁾					_	ANSELE2	ANSELE1	ANSELE0	200
INLVLE	_	_	_	_	INLVLE3	INLVLE2 ⁽²⁾	INLVLE1 ⁽²⁾	INLVLE0 ⁽²⁾	202
LATE ⁽²⁾	_	_	_	_	_	LATE2	LATE1	LATE0	200
ODCONE ⁽²⁾	_	_	_	_	_	ODE2	ODE1	ODE0	201
PORTE	_	_	_	_	RE3	RE2 ⁽²⁾	RE1 ⁽²⁾	RE0 ⁽²⁾	199
SLRCONE ⁽²⁾	_	_	_	_	_	SLRE2	SLRE1	SLRE0	202
TRISE	-	-	-	-	(1)	TRISE2 ⁽²⁾	TRISE1 ⁽²⁾	TRISE0 ⁽²⁾	199
WPUE	_	_	_	_	WPUE3	WPUE2 ⁽²⁾	WPUE1(2)	WPUE0(2)	201

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by PORTE.

Note 1: Unimplemented, read as '1'.

2: PIC16(L)F1777/9 only.

12.0 PERIPHERAL PIN SELECT (PPS) MODULE

The Peripheral Pin Select (PPS) module connects peripheral inputs and outputs to the device I/O pins. Only digital signals are included in the selections. All analog inputs and outputs remain fixed to their assigned pins. Input and output selections are independent as shown in the simplified block diagram Figure 12-1.

12.1 PPS Inputs

Each peripheral has a PPS register with which the inputs to the peripheral are selected. Inputs include the device pins.

Multiple peripherals can operate from the same source simultaneously. Port reads always return the pin level regardless of peripheral PPS selection. If a pin also has associated analog functions, the ANSEL bit for that pin must be cleared to enable the digital input buffer.

Although every peripheral has its own PPS input selection register, the selections are identical for every peripheral as shown in Register 12-1.

Note: The notation "xxx" in the register name is a place holder for the peripheral identifier. For example, CLC1PPS.

12.2 PPS Outputs

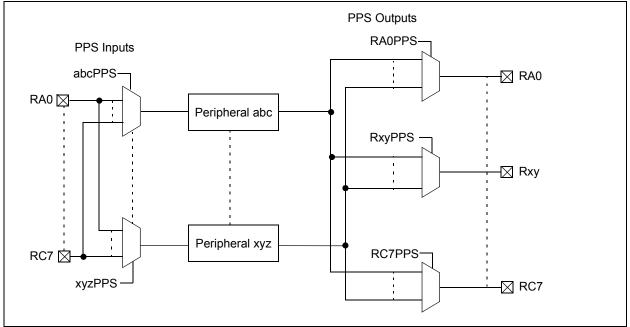
Each I/O pin has a PPS register with which the pin output source is selected. With few exceptions, the port TRIS control associated with that pin retains control over the pin output driver. Peripherals that control the pin output driver as part of the peripheral operation will override the TRIS control as needed. These peripherals include:

- · EUSART (synchronous operation)
- MSSP (I²C)
- · COG (auto-shutdown)

Although every pin has its own PPS peripheral selection register, the selections are identical for every pin as shown in Register 12-2.

Note: The notation "Rxy" is a place holder for the pin identifier. For example, RA0PPS.

FIGURE 12-1: SIMPLIFIED PPS BLOCK DIAGRAM



12.3 Bidirectional Pins

PPS selections for peripherals with bidirectional signals on a single pin must be made so that the PPS input and PPS output select the same pin. Peripherals that have bidirectional signals include:

- EUSART (synchronous operation)
- MSSP (I²C)

Note: The I²C default input pins are I²C and SMBus compatible and are the only pins on the device with this compatibility.

12.4 PPS Lock

The PPS includes a mode in which all input and output selections can be locked to prevent inadvertent changes. PPS selections are locked by setting the PPSLOCKED bit of the PPSLOCK register. Setting and clearing this bit requires a special sequence as an extra precaution against inadvertent changes. Examples of setting and clearing the PPSLOCKED bit are shown in Example 12-1.

EXAMPLE 12-1: PPS LOCK/UNLOCK SEQUENCE

; suspend interrupts
bcf INTCON,GIE
; BANKSEL PPSLOCK ; set bank
; required sequence, next 5 instructions
movlw 0x55
movwf PPSLOCK
movlw 0xAA
movwf PPSLOCK
; Set PPSLOCKED bit to disable writes or
; Clear PPSLOCKED bit to enable writes
bsf PPSLOCK, PPSLOCKED
; restore interrupts
bsf INTCON,GIE
1

12.5 PPS Permanent Lock

The PPS can be permanently locked by setting the PPS1WAY Configuration bit. When this bit is set, the PPSLOCKED bit can only be cleared and set one time after a device Reset. This allows for clearing the PPSLOCKED bit so that the input and output selections can be made during initialization. When the PPSLOCKED bit is set after all selections have been made, it will remain set and cannot be cleared until after the next device Reset event.

12.6 Operation During Sleep

PPS input and output selections are unaffected by Sleep.

12.7 Effects of a Reset

A device Power-on Reset (POR) clears all PPS input and output selections to their default values. All other Resets leave the selections unchanged. Default input selections are shown in Table 12-1.

12.8 Register Definitions: PPS Input Selection

REGISTER 12-1: xxxPPS: PERIPHERAL xxx INPUT SELECTION

U-0	U-0	R/W-q/u	R/W-q/u	R/W-q/u	R/W-q/u	R/W-q/u	R/W-q/u				
_	—			XXXPF	PS<5:0>						
bit 7							bit 0				
Legend:											
R = Readable	bit	W = Writable	bit	U = Unimplen	nented bit, read	l as '0'					
u = Bit is unch	anged	x = Bit is unkr	nown	-n/n = Value a	at POR and BO	R/Value at all o	ther Resets				
'1' = Bit is set		'0' = Bit is cle	ared	red q = value depends on peripheral							

bit 7-6	Unimplemented: Read as '0'
bit 5-3	<pre>xxxPPS<5:3>: Peripheral xxx Input PORT Selection bits 100 = Peripheral input is PORTE 011 = Peripheral input is PORTD⁽¹⁾ 010 = Peripheral input is PORTC 001 = Peripheral input is PORTB 000 = Peripheral input is PORTA</pre>
bit 2-0	<pre>xxxPPS<2:0>: Peripheral xxx Input Bit Selection bits⁽¹⁾ 111 = Peripheral input is from PORTx Bit 7 (Rx7) 110 = Peripheral input is from PORTx Bit 6 (Rx6) 101 = Peripheral input is from PORTx Bit 5 (Rx5) 100 = Peripheral input is from PORTx Bit 4 (Rx4) 011 = Peripheral input is from PORTx Bit 3 (Rx3) 010 = Peripheral input is from PORTx Bit 2 (Rx2) 001 = Peripheral input is from PORTx Bit 1 (Rx1) 000 = Peripheral input is from PORTx Bit 0 (Rx0)</pre>

- **Note 1:** See Table 12-1 for xxxPPS register list and Reset values.
 - 2: PIC16(L)F1777/9 only.

REGISTER 12-2: RxyPPS: PIN Rxy OUTPUT SOURCE SELECTION REGISTER

U-0	U-0	R/W-0/u	R/W-0/u	R/W-0/u	R/W-0/u	R/W-0/u	R/W-0/u		
—				RxyPF	PS<5:0>				
bit 7		·					bit 0		
Legend:									
R = Readable	bit	W = Writable b	W = Writable bit U = Unimplemented bit, read as '0'						
u = Bit is uncha	anged	x = Bit is unkne	x = Bit is unknown -n/n = Value at POR and BOR/Value at all other Resets						

bit 7-6 Unimplemented: Read as '0'

'1' = Bit is set

bit 5-0 **RxyPPS<5:0>:** Pin Rxy Output Source Selection bits Selection code determines the output signal on the port pin. See Table 12-2 for the selection codes

'0' = Bit is cleared

REGISTER 12-3: PPSLOCK: PPS LOCK REGISTER

U-0	U-0	U-0	U-0	U-0	U-0	U-0	R/W-0/0			
—	—	—	—	—	_	_	PPSLOCKED			
bit 7							bit 0			
Legend:										
R = Readable	bit	W = Writable I	oit	U = Unimplemented bit, read as '0'						
u = Bit is unch	anged	x = Bit is unkn	own	-n/n = Value at POR and BOR/Value at all other Resets						
'1' = Bit is set		'0' = Bit is clea	ared							

bit 7-1 Unimplemented: Read as '0'

bit 0 PPSLOCKED: PPS Locked bit

1 = PPS is locked. PPS selections can not be changed.

0 = PPS is not locked. PPS selections can be changed.

Peripheral	xxxPPS	Default Pin Selection	Reset Value (xxxPPS<5:0>)			t Selec 6(L)F1				tion 1778	
-	Register	PIC16(L)F1777/8/9	PIC16(L)F1777/8/9	Α	В	С	D	Е	Α	В	С
Interrupt-on-change	INTPPS	RB0	001000	•	•				•	٠	
Timer0clock	T0CKIPPS	RA4	000100	٠	•				•	٠	
Timer1clock	T1CKIPPS	RC0	010000	٠		•			•		٠
Timer1 gate	T1GPPS	RB5	001101		•	•				•	٠
Timer3 clock	T3CKIPPS	RC0	010000		•	٠				•	٠
Timer3 gate	T3GPPS	RC0	010000	•		٠			•		٠
Timer5 clock	T5CKIPPS	RC2	010010	•		•			•		•
Timer5 gate	T5GPPS	RB4	001100		•		•			٠	•
Timer2 input	T2INPPS	RC3	010011	•		٠			•		•
Timer4 input	T4INPPS	RC5	010101		•	•				•	•
Timer6 input	T6INPPS	RB7	001111		•	1	•		İ	٠	•
Timer8 input	T8INPPS	RC4	010100		•	ĺ	•		1	•	•
CCP1	CCP1PPS	RC2	010010		•	•	1		İ	•	•
CCP2	CCP2PPS	RC1	010001		•	•	1		İ	٠	•
CCP7	CCP7PPS	RB5	001101		•	1	•		İ	٠	•
CCP8 ⁽¹⁾	CCP8PPS	RB0	001000		•		•			٠	•
COG1	COG1INPPS	RB0	001000		•		•			٠	•
COG2	COG2INPPS	RB1	001001		•		•			•	•
COG3	COG3INPPS	RB2	001010		•		•			•	•
COG4 ⁽¹⁾	COG4INPPS	RB3	001011		•		•				
DSM1 low carrier	MD1CLPPS	RA3	000011	•			•		•		•
DSM1 high carrier	MD1CHPPS	RA4	000100	•			•		•		•
DSM1 modulation	MD1MODPPS	RA5	000101	•			•		•		•
DSM2 low carrier	MD2CLPPS	RC3	010011	•			•		•		•
DSM2 high carrier	MD2CHPPS	RC4	010100	•			•		•		•
DSM2 modulation	MD2MODPPS	RC5	010101	•			•		•		•
DSM3 low carrier	MD3CLPPS	RB3	001011		•		•			٠	•
DSM3 high carrier	MD3CHPPS	RB4	001100		•		•			•	•
DSM3 modulation	MD3MODPPS	RB5	001101		•		•			٠	•
DSM4 low carrier ⁽¹⁾	MD4CLPPS	RB0	001000		•		•				
DSM4 high carrier ⁽¹⁾	MD4CHPPS	RB1	001001		•		•				
DSM4 modulation ⁽¹⁾	MD4MODPPS	RB2	001010		•		•				<u> </u>
PRG1 set rising	PRG1RPPS	RA4	000100	٠		1	•		•		•
PRG1 set falling	PRG1FPPS	RA5	000101	•			•		•		•
PRG2 set rising	PRG2RPPS	RC1	010001	•	t		•		•		•
PRG2 set falling	PRG2FPPS	RC2	010010	٠		1	•		•		•
PRG3 set rising	PRG3RPPS	RC4	010100		•		•			٠	•
PRG3 set falling	PRG3FPPS	RC5	010101		•		•			•	•
PRG4 set rising ⁽¹⁾	PRG4RPPS	RB1	010100		•		•				
PRG4set falling ⁽¹⁾	PRG4FPPS	RB2	010101		•		•		1		
ADC trigger	ADCACTPPS	RB4	001100		•		•			•	•

TABLE 12-1: PPS INPUT REGISTER RESET VALUES

Example: CCP1PPS = 0x13 selects RC3 as the CCP1 input.

Note 1: PIC16(L)F1777/9 only

TABLE 12-1: PPS INPUT REGISTER RESET VALUES (CONTINUED)

Peripheral	xxxPPS	Default Pin Selection	Reset Value (xxxPPS<5:0>)	Port Selection PIC16(L)F1777/9					Port Selection PIC16(L)F177		
•	Register	PIC16(L)F1777/8/9	PIC16(L)F1777/8/9	Α	В	С	D	Е	Α	В	С
SPI and I ² C clock	SSPCLKPPS	RC3	010011		٠	•				٠	٠
SPI and I ² C data	SSPDATPPS	RC4	010100		•	•				٠	٠
SPI slave select	SSPSSPPS	RA5	000101	٠			•		•		٠
EUSART RX	RXPPS	RC7	010111		•	•				٠	٠
EUSART CK	CKPPS	RC6	010110		•	٠				٠	•
All CLCs	CLCIN0PPS	RA0	000000	٠		•			•		٠
All CLCs	CLCIN1PPS	RA1	000001	٠		•			•		٠
All CLCs	CLCIN2PPS	RB6	001110		•		•			٠	٠
All CLCs	CLCIN3PPS	RB7	001111		٠		•			٠	٠

Example: CCP1PPS = 0x13 selects RC3 as the CCP1 input.

Note 1: PIC16(L)F1777/9 only

D DD0 5 0	Output Signal	PIC	C16(L)F1	778	PIC16(L)F1777/9						
RxyPPS<5:0>	Output Signal	Α	В	С	Α	В	С	D	Е		
110001	MD4_out ⁽³⁾				_	•	_	•			
110000	MD3_out		•	٠	_	٠	_	•	_		
101111	MD2_out	•		•	•		_	•	_		
101110	MD1_out	•		•	•		_	•	_		
101101	sync_C8OUT ⁽³⁾				_	•	_	•	_		
101100	sync_C7OUT ⁽³⁾				_	•	_	•			
101011	sync_C6OUT	•		•	•		_	_	•		
101010	sync_C5OUT	•		•	•		_	•	_		
101001	sync_C4OUT		•	•	_	•	_	•	_		
101000	sync_C3OUT		•	•	_	•	_	•			
100111	sync_C2OUT	•		•	•		_		•		
100110	sync_C1OUT	•		•	•		_	•	_		
100101	DT		•	•	_	•	•	_			
100100	TX/CK		•	•	_	•	•	_	_		
100011	SDO		•	•	_	•	•	_	_		
100010	SDA		•	•	_	•	•	_			
100001	SCK/SCL ⁽¹⁾		•	•	_	•	•	_	_		
100000	PWM12_out ⁽³⁾				_	•	_	•	_		
011111	PWM11_out		•	•	_	•	_	•			
011110	PWM6_out	•		•	•		_	•	_		
011101	PWM5_out	•		•	•			•	_		
011100	PWM10_out ⁽³⁾				•		•	_	_		
011011	PWM9_out	•		•	•	_	•		_		
011010	PWM4_out	•		•	•			•			
011001	PWM3_out	•		•	•			_	•		
011000	CCP8_out ⁽³⁾				_	•	_	•	_		
010111	CCP7_out		•	•	_	•	_	•	_		
010110	CCP2_out		•	•	_	•	•	_			
010101	CCP1_out		•	•	_	•	•	_	_		
010100	COG4D ^(1,3)				•			•	_		
010011	COG4C ^(1,3)				•			•			
010010	COG4B ^(1,3)				•			_	•		
010001	COG4A ^(1,3)					•	•	_			
010000	COG3D ⁽¹⁾	•		•	•		_	•			
001111	COG3C ⁽¹⁾	•		•	•		_	•			
001110	COG3B ⁽¹⁾	•		•	•		_	_	•		
001101	COG3A ⁽¹⁾	•	<u> </u>	•		•	•				
001100	COG2D ⁽¹⁾	-	•	•	_	•	_	•			
001011	COG2C ⁽¹⁾		•	•		•	_	•	_		
001010	COG2B ⁽¹⁾		•	•		•		•			

TABLE 12-2: AVAILABLE PORTS FOR OUTPUT BY PERIPHERAL⁽²⁾

Note 1: TRIS control is overridden by the peripheral as required.

2: Unsupported peripherals will output a '0'.

3: PIC16(L)F1777/9 only.

PIC16(L)F1777/8/9

TABLE 12-2: AVAILABLE PORTS FOR OUTPUT BY PERIPHERAL⁽²⁾ (CONTINUED)

	Output Cinnel	PIC	C16(L)F1	778	PIC16(L)F1777/9					
RxyPPS<5:0>	Output Signal	Α	В	С	Α	В	С	D	Е	
001001	COG2A ⁽¹⁾		•	•		•	•			
001000	COG1D ⁽¹⁾		•	•		•		•		
000111	COG1C ⁽¹⁾		•	•		•		•		
000110	COG1B ⁽¹⁾		•	•		•		•		
000101	COG1A ⁽¹⁾		•	•		•	•			
000100	LC4_out		•	•		•		•		
000011	LC3_out		•	•		•		•		
000010	LC2_out	•		•	•		•			
000001	LC1_out	•		•	•		•			
000000	LATxy	•	•	•	•	•	•	•	•	

Note 1: TRIS control is overridden by the peripheral as required.

2: Unsupported peripherals will output a '0'.

3: PIC16(L)F1777/9 only.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on page
PPSLOCK	_		_		_			PPSLOCKED	206
INTPPS	_	_			INTF	PS<5:0>			205
TOCKIPPS	_	_				PPS<5:0>			205
T1CKIPPS						PPS<5:0>			205
T1GPPS	_	_			T1GF	PS<5:0>			205
T3CKIPPS	_	_			T3CK	PPS<5:0>			205
T3GPPS	_	_			T3GF	PS<5:0>			205
T5CKIPPS	_	_			T5CK	PPS<5:0>			205
T5GPPS	_	_			T5GF	PS<5:0>			205
T2INPPS	_	_			T2IN	PPS<5:0>			205
T4INPPS	_	_			T4IN	PPS<5:0>			205
T6INPPS	_	_			T6IN	PS<5:0>			205
T8INPPS	_	_			T8IN	PS<5:0>			205
CCP1PPS	_	_			CCP1	PPS<5:0>			205
CCP2PPS	—	_			CCP2	PPS<5:0>			205
CCP7PPS	_	_			CCP7	PPS<5:0>			205
CCP8PPS ⁽¹⁾	_	_		CCP8PPS<5:0>					
COGIN1PPS	_	_		COG1PPS<5:0>					
COG2INPPS	_	_		COG2PPS<5:0>					
COG3INPPS	_	_		COG3PPS<5:0>					
COG4INPPS ⁽¹⁾	_				COG4	PPS<5:0>			205
MD1CLPPS	_	—			MD1CI	_PPS<5:0>			205
MD1CHPPS	_	_			MD1CI	HPPS<5:0>			205
MD1MODPPS	_	_			MD1MC	DPPS<5:0	>		205
MD2CLPPS	_	_			MD2CI	_PPS<5:0>			205
MD2CHPPS	_	_			MD2CI	HPPS<5:0>			205
MD2MODPPS	_	_			MD2MC	DPPS<5:0	>		205
MD3CLPPS	_	_			MD3CI	_PPS<5:0>			205
MD3CHPPS	_	_			MD3CI	HPPS<5:0>			205
MD3MODPPS	_	_			MD3MC	DPPS<5:0	>		205
MD4CLPPS ⁽¹⁾	_	_			MD4CI	_PPS<5:0>			205
MD4CHPPS ⁽¹⁾	_	_			MD4CI	HPPS<5:0>			205
MD4MODPPS ⁽¹⁾	_				MD4MC	DPPS<5:0	>		205
PRG1RPPS			PRG1RPPS<5:0>						205
PRG1FPPS	_	_	PRG1FPPS<5:0>						205
PRG2RPPS	_	_	PRG2RPPS<5:0>						205
PRG2FPPS	_	_	PRG2FPPS<5:0>						205
PRG3RPPS	_	_	PRG3RPPS<5:0>						205
PRG3FPPS	_	—		PRG3FPPS<5:0>					
PRG4RPPS	_	_				RPPS<5:0>			205
PRG4FPPS	_	_				PPS<5:0>			205
CLC1IN0PPS	_	_				0PPS<5:0>			205

TABLE 12-3: SUMMARY OF REGISTERS ASSOCIATED WITH THE PPS MODULE

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TABLE 12-3: SUMMARY OF REGISTERS ASSOCIATED WITH THE PPS MODULE (CONTINUED)

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on page
CLC1IN1PPS	_	—		CLCIN1PPS<5:0>					205
CLC1IN2PPS	_	—		CLCIN2PPS<5:0>					
CLC1IN3PPS	_	_		CLCIN3PPS<5:0>					205
ADCACTPPS	—	—		ADCACTPPS<5:0>					205
SSPCLKPPS	—	_		SSPCLKPPS<5:0>					205
SSPDATPPS	—	_		SSPDATPPS<5:0>					205
SSPSSPPS	—	_		SSPSSPPS<5:0>				205	
RXPPS	—	_		RXPPS<5:0>				205	
CKPPS	_	_		CKPPS<5:0>					

Legend: — = unimplemented, read as '0'. Shaded cells are unused by the PPS module.

Note 1: PIC16(L)F1777/9 only.

13.0 INTERRUPT-ON-CHANGE

All pins on all ports can be configured to operate as Interrupt-on-Change (IOC) pins. An interrupt can be generated by detecting a signal that has either a rising edge or a falling edge. Any individual pin, or combination of pins, can be configured to generate an interrupt. The interrupt-on-change module has the following features:

- Interrupt-on-Change enable (Master Switch)
- Individual pin configuration
- Rising and falling edge detection
- Individual pin interrupt flags

Figure 13-1 is a block diagram of the IOC module.

13.1 Enabling the Module

To allow individual pins to generate an interrupt, the IOCIE bit of the INTCON register must be set. If the IOCIE bit is disabled, the edge detection on the pin will still occur, but an interrupt will not be generated.

13.2 Individual Pin Configuration

For each pin, a rising edge detector and a falling edge detector are present. To enable a pin to detect a rising edge, the associated bit of the IOCxP register is set. To enable a pin to detect a falling edge, the associated bit of the IOCxN register is set.

A pin can be configured to detect rising and falling edges simultaneously by setting the associated bits in both of the IOCxP and IOCxN registers.

13.3 Interrupt Flags

The bits located in the IOCxF registers are status flags that correspond to the Interrupt-on-Change pins of each port. If an expected edge is detected on an appropriately enabled pin, then the status flag for that pin will be set, and an interrupt will be generated if the IOCIE bit is set. The IOCIF bit of the INTCON register reflects the status of all IOCxF bits.

13.4 Clearing Interrupt Flags

The individual status flags, (IOCxF register bits), can be cleared by resetting them to zero. If another edge is detected during this clearing operation, the associated status flag will be set at the end of the sequence, regardless of the value actually being written.

In order to ensure that no detected edge is lost while clearing flags, only AND operations masking out known changed bits should be performed. The following sequence is an example of what should be performed.

EXAMPLE 13-1: CLEARING INTERRUPT FLAGS (PORTA EXAMPLE)

MOVLW 0xff XORWF IOCAF, W ANDWF IOCAF, F

13.5 Operation in Sleep

The Interrupt-on-Change interrupt sequence will wake the device from Sleep mode, if the IOCIE bit is set.

If an edge is detected while in Sleep mode, the affected IOCxF register will be updated prior to the first instruction executed out of Sleep.

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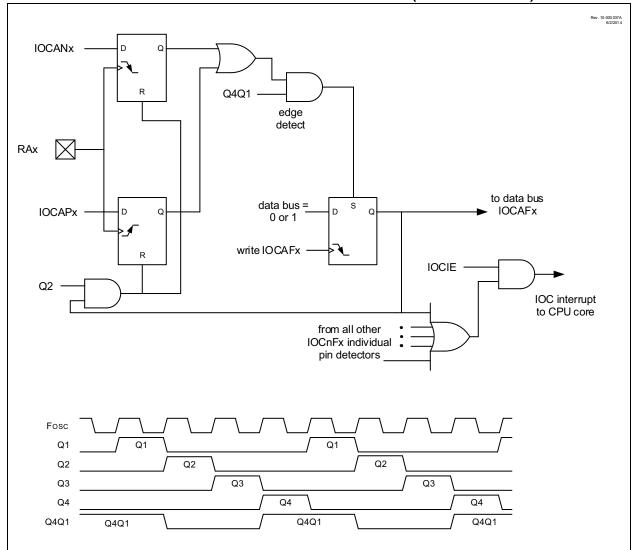


FIGURE 13-1: INTERRUPT-ON-CHANGE BLOCK DIAGRAM (PORTA EXAMPLE)

13.6 Register Definitions: Interrupt-on-Change Control

REGISTER 13-1: IOCAP: INTERRUPT-ON-CHANGE PORTA POSITIVE EDGE REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
IOCAP7	IOCAP6	IOCAP5	IOCAP4	IOCAP3	IOCAP2	IOCAP1	IOCAP0
bit 7	•						bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0

IOCAP<7:0>: Interrupt-on-Change PORTA Positive Edge Enable bits

- 1 = Interrupt-on-Change enabled on the pin for a positive going edge. IOCAFx bit and IOCIF flag will be set upon detecting an edge.
- 0 = Interrupt-on-Change disabled for the associated pin.

REGISTER 13-2: IOCAN: INTERRUPT-ON-CHANGE PORTA NEGATIVE EDGE REGISTER

IOCAN7 IOCAN6 IOCAN5 IOCAN4 IOCAN3 IOCAN2 IOCAN1 IOCAN0 bit 7 bit 0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
bit 7 bit 0	IOCAN7	IOCAN6	IOCAN5	IOCAN4	IOCAN3	IOCAN2	IOCAN1	IOCAN0
	bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0

IOCAN<7:0>: Interrupt-on-Change PORTA Negative Edge Enable bits

- 1 = Interrupt-on-Change enabled on the pin for a negative going edge. IOCAFx bit and IOCIF flag will be set upon detecting an edge.
- 0 = Interrupt-on-Change disabled for the associated pin.

R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0
IOCAF7	IOCAF6	IOCAF5	IOCAF4	IOCAF3	IOCAF2	IOCAF1	IOCAF0
bit 7							bit 0
Legend:							

REGISTER 13-3: IOCAF: INTERRUPT-ON-CHANGE PORTA FLAG REGISTER

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	HS - Bit is set in hardware

bit 7-0 IOCAF<7:0>: Interrupt-on-Change PORTA Flag bits

1 = An enabled change was detected on the associated pin.

Set when IOCAPx = 1 and a rising edge was detected on RAx, or when IOCANx = 1 and a falling edge was detected on RAx.

0 = No change was detected, or the user cleared the detected change.

REGISTER 13-4: IOCBP: INTERRUPT-ON-CHANGE PORTB POSITIVE EDGE REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
IOCBP7	IOCBP6	IOCBP5	IOCBP4	IOCBP3	IOCBP2	IOCBP1	IOCBP0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 IOCBP<7:0>: Interrupt-on-Change PORTB Positive Edge Enable bits

1 = Interrupt-on-Change enabled on the pin for a positive going edge. IOCBFx bit and IOCIF flag will be set upon detecting an edge.

0 = Interrupt-on-Change disabled for the associated pin.

REGISTER 13-5: IOCBN: INTERRUPT-ON-CHANGE PORTB NEGATIVE EDGE REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
IOCBN7	IOCBN6	IOCBN5	IOCBN4	IOCBN3	IOCBN2	IOCBN1	IOCBN0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0

IOCBN<7:0>: Interrupt-on-Change PORTB Negative Edge Enable bits

- 1 = Interrupt-on-Change enabled on the pin for a negative going edge. IOCBFx bit and IOCIF flag will be set upon detecting an edge.
- 0 = Interrupt-on-Change disabled for the associated pin.

R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0
IOCBF7	IOCBF6	IOCBF5	IOCBF4	IOCBF3	IOCBF2	IOCBF1	IOCBF0
bit 7							bit 0
Legend:							
R = Readable bit W = Writable bit		bit	U = Unimpler	nented bit, read	as '0'		
u = Bit is unchanged x = Bit is unkn		nown	-n/n = Value a	at POR and BO	R/Value at all o	ther Resets	
'1' = Bit is set '0' = Bit is cle		ared	HS - Bit is set	t in hardware			

REGISTER 13-6: IOCBF: INTERRUPT-ON-CHANGE PORTB FLAG REGISTER

bit 7-0 IOCBF<7:0>: Interrupt-on-Change PORTB Flag bits

1 = An enabled change was detected on the associated pin.

- Set when IOCBPx = 1 and a rising edge was detected on RBx, or when IOCBNx = 1 and a falling edge was detected on RBx.
- 0 = No change was detected, or the user cleared the detected change.

REGISTER 13-7: IOCCP: INTERRUPT-ON-CHANGE PORTC POSITIVE EDGE REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
IOCCP7	IOCCP6	IOCCP5	IOCCP4	IOCCP3	IOCCP2	IOCCP1	IOCCP0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0

0 **IOCCP<7:0>:** Interrupt-on-Change PORTC Positive Edge Enable bits

1 = Interrupt-on-Change enabled on the pin for a positive going edge. IOCCFx bit and IOCIF flag will be set upon detecting an edge.

0 = Interrupt-on-Change disabled for the associated pin.

REGISTER 13-8: IOCCN: INTERRUPT-ON-CHANGE PORTC NEGATIVE EDGE REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
IOCCN7	IOCCN6	IOCCN5	IOCCN4	IOCCN3	IOCCN2	IOCCN1	IOCCN0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0

IOCCN<7:0>: Interrupt-on-Change PORTC Negative Edge Enable bits

1 = Interrupt-on-Change enabled on the pin for a negative going edge. IOCCFx bit and IOCIF flag will be set upon detecting an edge.

0 = Interrupt-on-Change disabled for the associated pin.

REGISTER 13-9: IOCCF: INTERRUPT-ON-CHANGE PORTC FLAG REGISTER

	R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0
ſ	IOCCF7	IOCCF6	IOCCF5	IOCCF4	IOCCF3	IOCCF2	IOCCF1	IOCCF0
	bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	HS - Bit is set in hardware

bit 7-0

IOCCF<7:0>: Interrupt-on-Change PORTC Flag bits

1 = An enabled change was detected on the associated pin.

Set when IOCCPx = 1 and a rising edge was detected on RCx, or when IOCCNx = 1 and a falling edge was detected on RCx.

0 = No change was detected, or the user cleared the detected change.

REGISTER 13-10: IOCEP: INTERRUPT-ON-CHANGE PORTE POSITIVE EDGE REGISTER

U-0	U-0	U-0	U-0	R/W-0/0	U-0	U-0	U-0	
—	—	—	—	IOCEP3	—	—	—	
bit 7							bit 0	
Legend:								
R = Readable bit		W = Writable bit		U = Unimplemented bit, read as '0'				
u = Bit is unchanged		x = Bit is unknown		-n/n = Value at POR and BOR/Value at all other Resets				
'1' = Bit is set '0' = Bit is clea		ared						
bit 7-4	Unimplemen	ted: Read as '	כ'					
 bit 3 IOCEP3: Interrupt-on-Change PORTE Positive Edge Enable bits 1 = Interrupt-on-Change enabled on the pin for a positive going edge. IOCEFx bit and IOCIF flag will be set upon detecting an edge. 0 = Interrupt-on-Change disabled for the associated pin. 								
bit 2-0	2-0 Unimplemented: Read as '0'							

REGISTER 13-11: IOCEN: INTERRUPT-ON-CHANGE PORTE NEGATIVE EDGE REGISTER

U-0	U-0	U-0	U-0	R/W-0/0	U-0	U-0	U-0
—	_			IOCEN3			—
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-4	Unimplemented: Read as '0'
bit 3	 IOCEN3: Interrupt-on-Change PORTE Negative Edge Enable bits 1 = Interrupt-on-Change enabled on the pin for a negative going edge. IOCEFx bit and IOCIF flag will be set upon detecting an edge. 0 = Interrupt-on-Change disabled for the associated pin.
bit 2-0	Unimplemented: Read as '0'

U-0	U-0	U-0	U-0	R/W/HS-0/0	U-0	U-0	U-0
—	—	—	_	IOCEF3	—	—	—
bit 7							bit 0

Legena:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	HS - Bit is set in hardware

bit 7-4 Unimplemented: Read as '0'

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bit 3 IOCEF3: Interrupt-on-Change PORTE Flag bits

1 = An enabled change was detected on the associated pin.
 Set when IOCEPx = 1 and a rising edge was detected on REx, or when IOCENx = 1 and a falling edge was detected on REx.

0 = No change was detected, or the user cleared the detected change.

bit 2-0 Unimplemented: Read as '0'

TABLE 13-1: SUMMARY OF REGISTERS ASSOCIATED WITH INTERRUPT-ON-CHANGE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ANSELA	—	—	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0	177
ANSELB	—	—	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	182
ANSELC	ANSC7	ANSC6	ANSC5	ANSC4	ANSC3	ANSC2	—	—	187
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	132
IOCAF	IOCAF7	IOCAF6	IOCAF5	IOCAF4	IOCAF3	IOCAF2	IOCAF1	IOCAF0	216
IOCAN	IOCAN7	IOCAN6	IOCAN5	IOCAN4	IOCAN3	IOCAN2	IOCAN1	IOCAN0	215
IOCAP	IOCAP7	IOCAP6	IOCAP5	IOCAP4	IOCAP3	IOCAP2	IOCAP1	IOCAP0	215
IOCBF	IOCBF7	IOCBF6	IOCBF5	IOCBF4	IOCBF3	IOCBF2	IOCBF1	IOCBF0	217
IOCBN	IOCBN7	IOCBN6	IOCBN5	IOCBN4	IOCBN3	IOCBN2	IOCBN1	IOCBN0	216
IOCBP	IOCBP7	IOCBP6	IOCBP5	IOCBP4	IOCBP3	IOCBP2	IOCBP1	IOCBP0	216
IOCCF	IOCCF7	IOCCF6	IOCCF5	IOCCF4	IOCCF3	IOCCF2	IOCCF1	IOCCF0	218
IOCCN	IOCCN7	IOCCN6	IOCCN5	IOCCN4	IOCCN3	IOCCN2	IOCCN1	IOCCN0	218
IOCCP	IOCCP7	IOCCP6	IOCCP5	IOCCP4	IOCCP3	IOCCP2	IOCCP1	IOCCP0	218
IOCEP	_	—	_	_	IOCEP3	—	—	_	219
IOCEN	_	_	_	_	IOCEN3	_	_	_	219
IOCEF	_	_	_	_	IOCEF3	—	_	_	220
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	176
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	181
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	186
TRISE	—		_	_	(1)	TRISE2 ⁽²⁾	TRISE1(2)	TRISE0 ⁽²⁾	199

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by interrupt-on-change.

Note 1: Unimplemented, read as '1'.

2: Unimplemented on PIC16(L)F1778.

14.0 FIXED VOLTAGE REFERENCE (FVR)

The Fixed Voltage Reference, or FVR, is a stable voltage reference, independent of VDD, with 1.024V, 2.048V or 4.096V selectable output levels. The output of the FVR can be configured to supply a reference voltage to the following:

- · ADC input channel
- · ADC positive reference
- Comparator positive input
- Digital-to-Analog Converter (DAC)

The FVR can be enabled by setting the FVREN bit of the FVRCON register.

14.1 Independent Gain Amplifiers

The output of the FVR supplied to the ADC, Comparators, and DAC is routed through two independent programmable gain amplifiers. Each amplifier can be programmed for a gain of 1x, 2x or 4x, to produce the three possible voltage levels.

The ADFVR<1:0> bits of the FVRCON register are used to enable and configure the gain amplifier settings for the reference supplied to the ADC module. Reference **Section 16.0 "Analog-to-Digital Converter (ADC) Module"** for additional information.

The CDAFVR<1:0> bits of the FVRCON register are used to enable and configure the gain amplifier settings for the reference supplied to the DAC and comparator module. Reference **Section 17.0 "5-Bit Digital-to-Analog Converter (DAC) Module**" and **Section 19.0 "Comparator Module**" for additional information.

14.2 FVR Stabilization Period

When the Fixed Voltage Reference module is enabled, it requires time for the reference and amplifier circuits to stabilize. Once the circuits stabilize and are ready for use, the FVRRDY bit of the FVRCON register will be set. See Figure 37-77: FVR Stabilization Period, PIC16LF1777/8/9 Only.

14.3 FVR Buffer Stabilization Period

When either FVR Buffer1 or Buffer2 is enabled then the buffer amplifier circuits require 30 μ s to stabilize.This stabilization time is required even when the FVR is already operating and stable.

PIC16(L)F1777/8/9



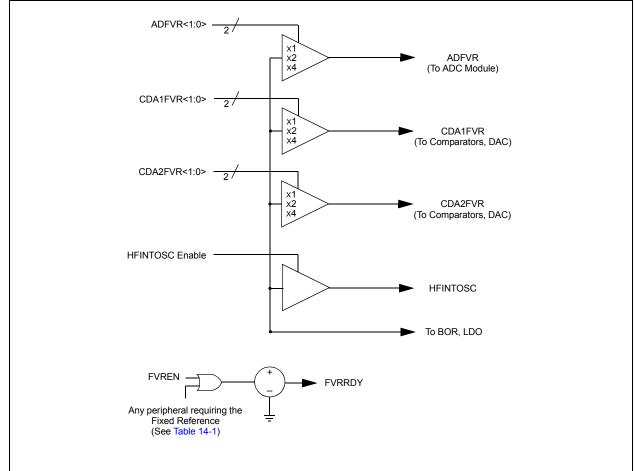


TABLE 14-1: PERIPHERALS REQUIRING THE FIXED VOLTAGE REFERENCE (FVR)

Peripheral	Conditions	Description
HFINTOSC	FOSC<2:0> = 100 and IRCF<3:0> ≠ 000x	INTOSC is active and device is not in Sleep
	BOREN<1:0> = 11	BOR always enabled
BOR	BOREN<1:0> = 10 and BORFS = 1	BOR disabled in Sleep mode, BOR Fast Start enabled
	BOREN<1:0> = 01 and BORFS = 1	BOR under software control, BOR Fast Start enabled
LDO	All PIC16F1773/6 devices, when VREGPM = 1 and not in Sleep	The device runs off of the ULP regulator when in Sleep mode

14.4 Register Definitions: FVR Control

REGISTER 14-1: FVRCON: FIXED VOLTAGE REFERENCE CONTROL REGISTER

R/W-0/0	R-q/q	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
FVREN	FVRRDY ⁽¹⁾	TSEN ⁽³⁾	TSRNG ⁽³⁾	CDAF\	/R<1:0>	ADFVI	R<1:0>
bit 7							bit 0

Legend:							
R = Reada	able bit	W = Writable bit	U = Unimplemented bit, read as '0'				
u = Bit is u	inchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets				
'1' = Bit is	set	'0' = Bit is cleared	q = Value depends on condition				
bit 7	1 = Fixed	Fixed Voltage Reference Ena I Voltage Reference is enable I Voltage Reference is disable	ed				
bit 6	1 = Fixed	FVRRDY: Fixed Voltage Reference Ready Flag bit ⁽¹⁾ 1 = Fixed Voltage Reference output is ready for use 0 = Fixed Voltage Reference output is not ready or not enabled					
bit 5	1 = Temp	TSEN: Temperature Indicator Enable bit ⁽³⁾ 1 = Temperature Indicator is enabled 0 = Temperature Indicator is disabled					
bit 4	1 = Vout	TSRNG: Temperature Indicator Range Selection bit ⁽³⁾ 1 = VOUT = VDD - 4VT (High Range) 0 = VOUT = VDD - 2VT (Low Range)					
bit 3-2	11 = Com 10 = Com 01 = Com	parator/DAC FVR Buffer Gai	in is 4x, with output VCDAFVR = 4x VFVR ⁽²⁾ in is 2x, with output VCDAFVR = 2x VFVR ⁽²⁾ in is 1x, with output VCDAFVR = 1x VFVR				
bit 1-0	11 = ADC 10 = ADC 01 = ADC	E: O>: ADC FVR Buffer Gain S FVR Buffer Gain is 4x, with FVR Buffer Gain is 2x, with FVR Buffer Gain is 1x, with FVR Buffer is off	output VADFVR = 4x VFVR ⁽²⁾ output VADFVR = 2x VFVR ⁽²⁾				
		vays '1' on PIC16F1773/6 or Reference output cannot exc	-				

- 2: Fixed Voltage Reference output cannot exceed VDD.
- 3: See Section 15.0 "Temperature Indicator Module" for additional information.

TABLE 14-2: SUMMARY OF REGISTERS ASSOCIATED WITH FIXED VOLTAGE REFERENCE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on page
FVRCON	FVREN	FVRRDY	TSEN	TSRNG	CDAFVR<1:0>		ADFVF	२<1:0>	223

Legend: Shaded cells are not used with the Fixed Voltage Reference.

15.0 TEMPERATURE INDICATOR MODULE

This family of devices is equipped with a temperature circuit designed to measure the operating temperature of the silicon die. The circuit's range of operating temperature falls between -40°C and +85°C. The output is a voltage that is proportional to the device temperature. The output of the temperature indicator is internally connected to the device ADC.

The circuit may be used as a temperature threshold detector or a more accurate temperature indicator, depending on the level of calibration performed. A one-point calibration allows the circuit to indicate a temperature closely surrounding that point. A two-point calibration allows the circuit to sense the entire range of temperature more accurately. Reference Application Note AN1333, *"Use and Calibration of the Internal Temperature Indicator"* (DS01333) for more details regarding the calibration process.

15.1 Circuit Operation

Figure 15-1 shows a simplified block diagram of the temperature circuit. The proportional voltage output is achieved by measuring the forward voltage drop across multiple silicon junctions.

Equation 15-1 describes the output characteristics of the temperature indicator.

EQUATION 15-1: VOUT RANGES

High Range: VOUT = VDD - 4VT

Low Range: VOUT = VDD - 2VT

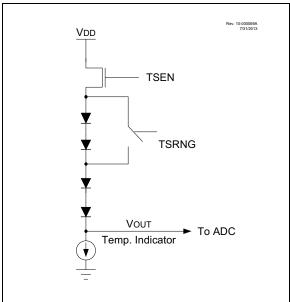
The temperature sense circuit is integrated with the Fixed Voltage Reference (FVR) module. See **Section 14.0 "Fixed Voltage Reference (FVR)**" for more information.

The circuit is enabled by setting the TSEN bit of the FVRCON register. When disabled, the circuit draws no current.

The circuit operates in either high or low range. The high range, selected by setting the TSRNG bit of the FVRCON register, provides a wider output voltage. This provides more resolution over the temperature range, but may be less consistent from part to part. This range requires a higher bias voltage to operate and thus, a higher VDD is needed.

The low range is selected by clearing the TSRNG bit of the FVRCON register. The low range generates a lower voltage drop and thus, a lower bias voltage is needed to operate the circuit. The low range is provided for low voltage operation.

FIGURE 15-1: TEMPERATURE CIRCUIT DIAGRAM



15.2 Minimum Operating VDD

When the temperature circuit is operated in low range, the device may be operated at any operating voltage that is within specifications.

When the temperature circuit is operated in high range, the device operating voltage, VDD, must be high enough to ensure that the temperature circuit is correctly biased.

Table 15-1 shows the recommended minimum VDD vs. range setting.

TABLE 15-1: RECOMMENDED VDD VS. RANGE

Min. Vdd, TSRNG = 1	Min. VDD, TSRNG = 0
3.6V	1.8V

15.3 Temperature Output

The output of the circuit is measured using the internal Analog-to-Digital Converter. A channel is reserved for the temperature circuit output. Refer to **Section 16.0 "Analog-to-Digital Converter (ADC) Module"** for detailed information.

15.4 ADC Acquisition Time

To ensure accurate temperature measurements, the user must wait at least 200 μ s after the ADC input multiplexer is connected to the temperature indicator output before the conversion is performed. In addition, the user must wait 200 μ s between sequential conversions of the temperature indicator output.

TABLE 15-2: SUMMARY OF REGISTERS ASSOCIATED WITH THE TEMPERATURE INDICATOR

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on page
FVRCON	FVREN	FVRRDY	TSEN	TSRNG	CDFVR<1:0>		ADFVF	R<1:0>	223

Legend: Shaded cells are unused by the temperature indicator module.

16.0 ANALOG-TO-DIGITAL CONVERTER (ADC) MODULE

The Analog-to-Digital Converter (ADC) allows conversion of an analog input signal to a 10-bit binary representation of that signal. This device uses analog inputs, which are multiplexed into a single sample and hold circuit. The output of the sample and hold is connected to the input of the converter. The converter generates a 10-bit binary result via successive approximation and stores the conversion result into the ADC result registers (ADRESH:ADRESL register pair). Figure 16-1 shows the block diagram of the ADC.

The ADC voltage reference is software selectable to be either internally generated or externally supplied.

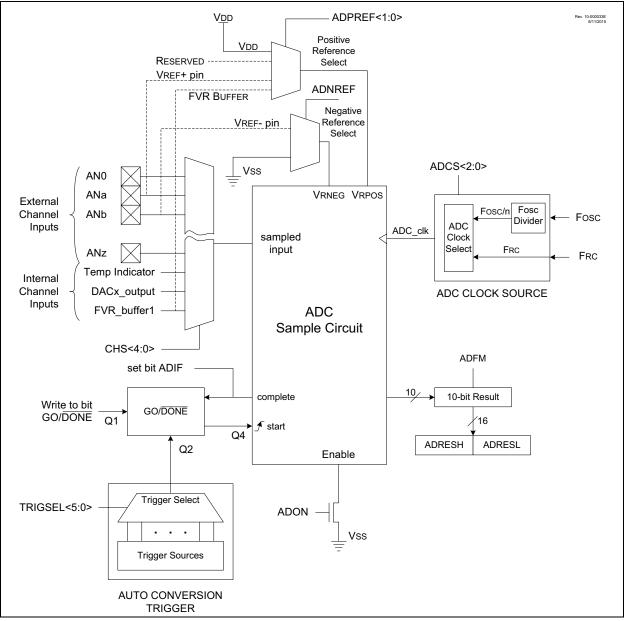


FIGURE 16-1: ADC BLOCK DIAGRAM

The ADC can generate an interrupt upon completion of a conversion. This interrupt can be used to wake-up the device from Sleep.

16.1 ADC Configuration

When configuring and using the ADC the following functions must be considered:

- Port configuration
- · Channel selection
- ADC voltage reference selection
- ADC conversion clock source
- Interrupt control
- Result formatting

16.1.1 PORT CONFIGURATION

The ADC can be used to convert both analog and digital signals. When converting analog signals, the I/O pin should be configured for analog by setting the associated TRIS and ANSEL bits. Refer to **Section 11.0 "I/O Ports"** for more information.

Note:	Analog voltages on any pin that is defined
	as a digital input may cause the input
	buffer to conduct excess current.

16.1.2 CHANNEL SELECTION

There are up to 27 channel selections available:

- AN<4:0> pins (PIC16(L)F1778 only)
- AN<11:8> pins (PIC16(L)F1778 only)
- AN<27:0> pins (PIC16(L)F1777/9 only)
- Temperature Indicator
- DAC1_output and DAC3_output
- DAC2_output and DAC4_output (PIC16(L)F1777/9 only)
- FVR_buffer1

The CHS bits of the ADCON0 register (Register 16-1) determine which channel is connected to the sample and hold circuit.

When changing channels, a delay is required before starting the next conversion. Refer to **Section 16.2 "ADC Operation**" for more information.

16.1.3 ADC POSITIVE VOLTAGE REFERENCE

The ADPREF bits of the ADCON1 register provide control of the positive voltage reference. The positive voltage reference can be:

- VREF+ pin
- VDD
- FVR 2.048V
- FVR 4.096V (Not available on LF devices)
- Vss

See Section 16.0 "Analog-to-Digital Converter (ADC) Module" for more details on the Fixed Voltage Reference.

16.1.4 ADC NEGATIVE VOLTAGE REFERENCE

The ADNREF bit of the ADCON1 register provides control of the negative voltage reference. The negative voltage reference can be:

- VREF- pin
- Vss

16.1.5 CONVERSION CLOCK

The source of the conversion clock is software selectable via the ADCS bits of the ADCON1 register. There are seven possible clock options:

- Fosc/2
- Fosc/4
- Fosc/8
- Fosc/16
- Fosc/32
- Fosc/64
- FRC (internal RC oscillator)

The time to complete one bit conversion is defined as TAD. One full 10-bit conversion requires 11.5 TAD periods as shown in Figure 16-2.

For correct conversion, the appropriate TAD specification must be met. Refer to Table 36-16: ADC Conversion Requirements for more information. Table 16-1 gives examples of appropriate ADC clock selections.

Note: Unless using the FRC, any changes in the system clock frequency will change the ADC clock frequency, which may adversely affect the ADC result.

TABLE 16-1: ADC CLOCK PERIOD (TAD) Vs. DEVICE OPERATING FREQUENCIES

ADC Clock Period (TAD)		Device Frequency (Fosc)							
ADC Clock Source	ADCS<2:0>	32 MHz	20 MHz	16 MHz	8 MHz	4 MHz	1 MHz		
Fosc/2	000	62.5ns ⁽²⁾	100 ns ⁽²⁾	125 ns ⁽²⁾	250 ns ⁽²⁾	500 ns ⁽²⁾	2.0 μs		
Fosc/4	100	125 ns ⁽²⁾	200 ns ⁽²⁾	250 ns ⁽²⁾	500 ns ⁽²⁾	1.0 μs	4.0 μs		
Fosc/8	001	0.5 μs ⁽²⁾	400 ns ⁽²⁾	0.5 μs ⁽²⁾	1.0 μs	2.0 μs	8.0 μs ⁽³⁾		
Fosc/16	101	800 ns	800 ns	1.0 μs	2.0 μs	4.0 μs	16.0 μs ⁽³⁾		
Fosc/32	010	1.0 μs	1.6 μs	2.0 μs	4.0 μs	8.0 μs ⁽³⁾	32.0 μs ⁽²⁾		
Fosc/64	110	2.0 μs	3.2 μs	4.0 μs	8.0 μs ⁽³⁾	16.0 μs ⁽²⁾	64.0 μs ⁽²⁾		
FRC	x11	1.0-6.0 μs ^(1,4)	1.0-6.0 μs ^(1,4)	1.0-6.0 μs ^(1,4)	1.0-6.0 μs ^(1,4)	1.0-6.0 μs ^(1,4)	1.0-6.0 μs ^(1,4)		

Legend: Shaded cells are outside of recommended range.

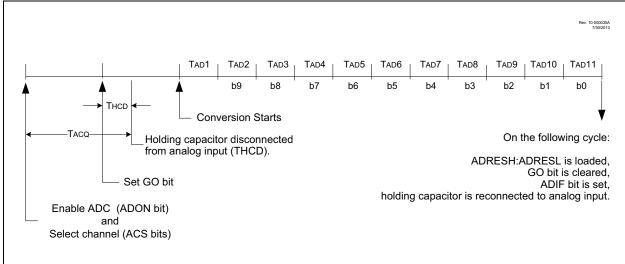
Note 1: See TAD parameter for FRC source typical TAD value.

2: These values violate the required TAD time.

3: Outside the recommended TAD time.

4: The ADC clock period (TAD) and total ADC conversion time can be minimized when the ADC clock is derived from the system clock FOSC. However, the FRC oscillator source must be used when conversions are to be performed with the device in Sleep mode.





16.1.6 INTERRUPTS

The ADC module allows for the ability to generate an interrupt upon completion of an Analog-to-Digital conversion. The ADC Interrupt Flag is the ADIF bit in the PIR1 register. The ADC Interrupt Enable is the ADIE bit in the PIE1 register. The ADIF bit must be cleared in software.

Note 1:	The ADIF bit is set at the completion of
	every conversion, regardless of whether
	or not the ADC interrupt is enabled.

2: The ADC operates during Sleep only when the FRC oscillator is selected.

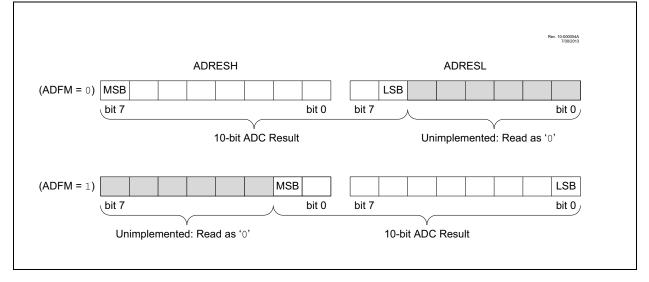
This interrupt can be generated while the device is operating or while in Sleep. If the device is in Sleep, the interrupt will wake-up the device. Upon waking from Sleep, the next instruction following the SLEEP instruction is always executed. If the user is attempting to wake-up from Sleep and resume in-line code execution, the ADIE bit of the PIE1 register and the PEIE bit of the INTCON register must both be set and the GIE bit of the INTCON register must be cleared. If all three of these bits are set, the execution will switch to the Interrupt Service Routine.

16.1.7 RESULT FORMATTING

The 10-bit ADC conversion result can be supplied in two formats, left justified or right justified. The ADFM bit of the ADCON1 register controls the output format.

Figure 16-3 shows the two output formats.

FIGURE 16-3: 10-BIT ADC CONVERSION RESULT FORMAT



16.2 ADC Operation

16.2.1 STARTING A CONVERSION

To enable the ADC module, the ADON bit of the ADCON0 register must be set to a '1'. Setting the GO/DONE bit of the ADCON0 register to a '1' will start the Analog-to-Digital conversion.

Note:	The GO/DONE bit should not be set in the
	same instruction that turns on the ADC.
	Refer to Section 16.2.6 "ADC Conver-
	sion Procedure".

16.2.2 COMPLETION OF A CONVERSION

When the conversion is complete, the ADC module will:

- Clear the GO/DONE bit
- · Set the ADIF Interrupt Flag bit
- Update the ADRESH and ADRESL registers with new conversion result

16.2.3 TERMINATING A CONVERSION

If a conversion must be terminated before completion, the GO/DONE bit can be cleared in software. The ADRESH and ADRESL registers will be updated with the partially complete Analog-to-Digital conversion sample. Incomplete bits will match the last bit converted.

Note:	A device Reset forces all registers to their
	Reset state. Thus, the ADC module is
	turned off and any pending conversion is
	terminated.

16.2.4 ADC OPERATION DURING SLEEP

The ADC module can operate during Sleep. This requires the ADC clock source to be set to the FRC option. When the FRC oscillator source is selected, the ADC waits one additional instruction before starting the conversion. This allows the SLEEP instruction to be executed, which can reduce system noise during the conversion. If the ADC interrupt is enabled, the device will wake-up from Sleep when the conversion completes. If the ADC interrupt is disabled, the ADC module is turned off after the conversion completes, although the ADON bit remains set.

When the ADC clock source is something other than FRC, a SLEEP instruction causes the present conversion to be aborted and the ADC module is turned off, although the ADON bit remains set.

16.2.5 AUTO-CONVERSION TRIGGER

The Auto-conversion Trigger allows periodic ADC measurements without software intervention. When a rising edge of the selected source occurs, the GO/DONE bit is set by hardware.

The Auto-conversion Trigger source is selected with the TRIGSEL<5:0> bits of the ADCON2 register.

Using the Auto-conversion Trigger does not assure proper ADC timing. It is the user's responsibility to ensure that the ADC timing requirements are met.

See Table 16-2 f	or auto-conversion sources.
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TABLE 16-2: AUTO-CONVERSION SOURCES

Source Peripheral	Signal Name
CCP1	CCP1_trigger
CCP2	CCP2_trigger
CCP7	CCP7_trigger
CCP8 ⁽¹⁾	CCP8_trigger
Timer0	T0_overflow
Timer1	T1_overflow
Timer3	T3_overflow
Timer5	T5_overflow
Timer2	T2_postscaled
Timer4	T4_postscaled
Timer6	T6_postscaled
Timer8	T8_postscaled
Comparator C1	sync_C1OUT
Comparator C2	sync_C2OUT
Comparator C3	sync_C3OUT
Comparator C4	sync_C4OUT
Comparator C5	sync_C5OUT
Comparator C6	sync_C6OUT
Comparator C7 ⁽¹⁾	sync_C7OUT
Comparator C8 ⁽¹⁾	sync_C8OUT
CLC1	LC1_out
CLC2	LC2_out
CLC3	LC3_out
CLC4	LC4_out
PWM3	PWM3OUT
PWM4	PWM4OUT
PWM9	PWM9OUT
PWM9	PR/PH/OF/DC9_match
PWM5	PR/PH/OF/DC5_match
PWM6	PR/PH/OF/DC6_match
PWM10 ⁽¹⁾	PR/PH/OF/DC10_match
PWM11	PR/PH/OF/DC11_match
PWM12 ⁽¹⁾	PR/PH/OF/DC12_match
ADCACT	ADCACTPPS Pin

Note 1: PIC16(L)F1777/9 only.

16.2.6 ADC CONVERSION PROCEDURE

This is an example procedure for using the ADC to perform an Analog-to-Digital conversion:

- 1. Configure Port:
 - Disable pin output driver (refer to the TRIS register)
 - Configure pin as analog (refer to the ANSEL register)
 - Disable weak pull-ups either globally (refer to the OPTION_REG register) or individually (refer to the appropriate WPUx register)
- 2. Configure the ADC module:
 - Select ADC conversion clock
 - · Configure voltage reference
 - · Select ADC input channel
 - Turn on ADC module
- 3. Configure ADC interrupt (optional):
 - · Clear ADC interrupt flag
 - Enable ADC interrupt
 - · Enable peripheral interrupt
 - Enable global interrupt⁽¹⁾
- 4. Wait the required acquisition time⁽²⁾.
- 5. Start conversion by setting the GO/DONE bit.
- 6. Wait for ADC conversion to complete by one of the following:
 - Polling the GO/DONE bit
 - Waiting for the ADC interrupt (interrupts enabled)
- 7. Read ADC Result.
- 8. Clear the ADC interrupt flag (required if interrupt is enabled).

Note 1: The global interrupt can be disabled if the user is attempting to wake-up from Sleep and resume in-line code execution.

2: Refer to Section 16.4 "ADC Acquisition Requirements".

EXAMPLE 16-1: ADC CONVERSION

;This code block configures the ADC ; for polling, Vdd and Vss references, FRC ;oscillator and ANO input. ;Conversion start & polling for completion ; are included. BANKSEL ADCON1 : B'11110000' ;Right justify, FRC MOVLW ;oscillator MOVWF ADCON1 ;Vdd and Vss Vref BANKSEL TRISA ; BSF TRISA,0 ;Set RA0 to input BANKSEL ANSEL ; BSF ANSEL,0 ;Set RA0 to analog BANKSEL WPUA BCF WPUA,0 ;Disable weak ;pull-up on RA0 BANKSEL ADCON0 B'00000001' ;Select channel AN0 MOVLW MOVWF ADCON0 ;Turn ADC On CALL SampleTime ;Acquisiton delay BSF ADCON0, ADGO ;Start conversion ADCON0, ADGO ; Is conversion done? BTFSC GOTO \$-1 ;No, test again BANKSEL ADRESH ; ;Read upper 2 bits ADRESH,W MOVF RESULTHI MOVWE ;store in GPR space BANKSEL ADRESL ; MOVF ADRESL,W ;Read lower 8 bits MOVWF RESULTLO ;Store in GPR space

16.3 Register Definitions: ADC Control

REGISTER 16-1: ADCON0: ADC CONTROL REGISTER 0

<pre>11111 = FVR (Fixed Voltage Reference) Buffer 1 Output⁽²⁾ 111101 = Temperature Indicator⁽³⁾ 111101 = DAC1_output⁽⁴⁾ 111011 = DAC3_output⁽⁴⁾ 110111 = DAC3_output⁽⁴⁾ 110111 = DAC3_output⁽⁴⁾ 110111 = DAC3_output⁽⁴⁾ 110111 = DAC3_output⁽⁴⁾ 110111 = DAC3_output⁽⁴⁾ 110111 = DAC3_output⁽⁴⁾ 110111 = DAC3_output⁽⁴⁾ 110111 = DAC3_output⁽⁴⁾ 110111 = DAC3_output⁽⁴⁾ 110111 = DAC3_output⁽⁴⁾ 110111 = Reserved. No channel connected. 110011 = Switched AN10⁽⁶⁾ 101011 = Reserved. No channel connected. 101011 = Reserved. No channel connected. 101011 = Reserved. No channel connected. 100011 = Reserved. No channel connected. 100011 = Reserved. No channel connected. 100011 = Reserved. No channel connected. 101011 = Reserved. No channel connected. 101011 = AN27⁽⁶⁾ 011010 = AN26⁽⁶⁾ 01101 = AN26⁽⁶⁾ 01101 = AN22⁽⁶⁾ 01001 = AN21⁽⁶⁾ 01001 = AN11⁽⁶⁾ 01001 = AN11⁽</pre>	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
Legend: R = Readable bit W = Writable bit U = Unimplemented bit, read as '0' = Bit is unchanged x = Bit is unknown -n/n = Value at POR and BOR/Value at all other Resets 1' = Bit is set '0' = Bit is cleared pt 7-2 CHS-5:0-: Analog Channel Select bits 111111 = FVR (Fixed Voltage Reference) Buffer 1 Output ⁽²⁾ 111110 = Temperature Indicator ⁽³⁾ 111101 = Tomperature Indicator ⁽³⁾ 111010 = DAC2_output ⁽⁴⁾ 111010 = DAC3_output ⁽⁴⁾ 111010 = DAC5_output ⁽⁴⁾ 110111 = DAC5_output ⁽⁴⁾ 110111 = Reserved. No channel connected. 110111 = Reserved. No channel connected. 110111 = Reserved. No channel connected. 110111 = Reserved. No channel connected. 110101 = Reserved. No channel connected. 110101 = Reserved. No channel connected. 110101 = Reserved. No channel connected. 110101 = Reserved. No channel connected. 110101 = Reserved. No channel connected. 110101 = Reserved. No channel connected. 110101 = Reserved. No channel connected. 110101 = Reserved. No channel connected. 110101 = Reserved. No channel connected. 110101 = Reserved. No channel connected. 110101 = Reserved. No channel connected. 110101 = Reserved. No channel connected. 110101 = Reserved. No channel connected. 110101 = Reserved. No channel connected. 110101 = Reserved. No channel connected. 110101 = Reserved. No channel connected. 110101 = Reserved. No channel connected. 110101 = Reserved. No channel connected. 110111 = RAZ6(0) 11010 = RAZ6(0) 11010 = RAZ6(0) 11010 = RAZ6(0			CHS<	5:0>			GO/DONE	ADON
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REGISTER 16-1: ADCON0: ADC CONTROL REGISTER 0 (CONTINUED)

bit 1	 GO/DONE: ADC Conversion Status bit 1 = ADC conversion cycle in progress. Setting this bit starts an ADC conversion cycle. This bit is automatically cleared by hardware when the ADC conversion has completed. 0 = ADC conversion completed/not in progress
bit 0	ADON: ADC Enable bit

1 = ADC is enabled

0 = ADC is disabled and consumes no operating current

See Section 17.0 "5-Bit Digital-to-Analog Converter (DAC) Module" for more information. See Section 14.0 "Fixed Voltage Reference (FVR)" for more information. See Section 15.0 "Temperature Indicator Module" for more information. Note 1: 2:

- 3:
- 4: 5:
- See Section 18.0 "10-bit Digital-to-Analog Converter (DAC) Module" for more information. Input source is switched off when op amp override is forcing tri-state. See Section 29.3 "Override Control" PIC16(L)F1777/9 only.
- 6:

PIC16(L)F1777/8/9

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	U-0	R/W-0/0	R/W-0/0	R/W-0/0
ADFM		ADCS<2:0>		_	ADNREF	ADPRE	EF<1:0>
bit 7					•		bit
Legend:							
R = Readab	ole bit	W = Writable I	bit	U = Unimpler	mented bit, read	l as '0'	
u = Bit is un	changed	x = Bit is unkn	iown	-n/n = Value a	at POR and BO	R/Value at all	other Resets
'1' = Bit is se	et	'0' = Bit is clea	ared				
bit 7	1 = Right ju loaded.	Result Format stified. Six Most	Significant b				
bit 6-4	111 = FRC 110 = Fosc 101 = Fosc 100 = Fosc	/16 /4 (clock supplied f /32 /8	from an intern	al RC oscillator			
bit 3	Unimpleme	nted: Read as '	o'				
bit 2	ADNREF: ADC Negative Voltage Reference Configuration bit 1 = VREF- is connected to external VREF- pin 0 = VREF- is connected to Vss						
bit 1-0	11 = VREF+	0>: ADC Positiv is connected to is connected to ved	internal Fixed	Voltage Refere		dule ⁽¹⁾	

Note 1: When selecting the VREF+ pin as the source of the positive reference, be aware that a minimum voltage specification exists. See Table 36-16: ADC Conversion Requirements for details.

REGISTER 16-3: ADCON2: ADC CONTROL REGISTER 2

	ged Unimplemen TRIGSEL<5:0 101101 = PW 10100 = PW 10101 = PW 10101 = PW 10100 = PW 100101 = PW 100101 = PW 100101 = PW 100101 = PW 100101 = PW 100101 = PW 100101 = PW 100101 = PW 100101 = PW 100010 = PW 100010 = PW 100011 = PW 100011 = PW 100010 = PW 100010 = PW 100010 = PW	W = Writable bit x = Bit is unknow '0' = Bit is cleare ted: Read as '0' D>: Auto-Conversio VM12 – OF12_mato VM12 – PH12_mato VM12 – DC12_mato VM12 – DC12_mato VM11 – OF11_mato VM11 – PH11_mato VM11 – PH11_mato VM11 – DC11_mato VM11 – OF6_match VM6 – PH6_match VM6 – DC6_match	vn n Trigger Sele ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾	U = Unimpleme -n/n = Value at	L<5:0> ⁽¹⁾ ented bit, read as POR and BOR/\	Resets
egend: = Readable bit = Bit is unchang ' = Bit is set it 7-6 it 5-0	ged Unimplemen TRIGSEL<5:0 101101 = PW 10100 = PW 10101 = PW 10101 = PW 10100 = PW 100101 = PW 100101 = PW 100101 = PW 100101 = PW 100101 = PW 100101 = PW 100101 = PW 100101 = PW 100101 = PW 100010 = PW 100010 = PW 100011 = PW 100011 = PW 100010 = PW 100010 = PW 100010 = PW	x = Bit is unknov '0' = Bit is cleare ted: Read as '0' D>: Auto-Conversio /M12 – OF12_matc /M12 – PR12_matc /M12 – PR12_matc /M12 – DC12_matc /M11 – OF11_matc /M11 – PH11_matc /M11 – PR11_matc /M11 – DC11_matc /M11 – DC11_matc /M6 – OF6_match /M6 – PR6_match	vn n Trigger Sele ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾	-n/n = Value at	-	
e = Readable bit = Bit is unchang ' = Bit is set it 7-6 it 5-0	ged Unimplemen TRIGSEL<5:0	x = Bit is unknov '0' = Bit is cleare ted: Read as '0' D>: Auto-Conversio /M12 – OF12_matc /M12 – PR12_matc /M12 – PR12_matc /M12 – DC12_matc /M11 – OF11_matc /M11 – PH11_matc /M11 – PR11_matc /M11 – DC11_matc /M11 – DC11_matc /M6 – OF6_match /M6 – PR6_match	vn n Trigger Sele ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾	-n/n = Value at	-	Resets
e = Readable bit = Bit is unchang ' = Bit is set it 7-6 it 5-0	ged Unimplemen TRIGSEL<5:0	x = Bit is unknov '0' = Bit is cleare ted: Read as '0' D>: Auto-Conversio /M12 – OF12_matc /M12 – PR12_matc /M12 – PR12_matc /M12 – DC12_matc /M11 – OF11_matc /M11 – PH11_matc /M11 – PR11_matc /M11 – DC11_matc /M11 – DC11_matc /M6 – OF6_match /M6 – PR6_match	vn n Trigger Sele ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾	-n/n = Value at	-	Resets
= Bit is unchang ' = Bit is set it 7-6 it 5-0	ged Unimplemen TRIGSEL<5:0	x = Bit is unknov '0' = Bit is cleare ted: Read as '0' D>: Auto-Conversio /M12 – OF12_matc /M12 – PR12_matc /M12 – PR12_matc /M12 – DC12_matc /M11 – OF11_matc /M11 – PH11_matc /M11 – PR11_matc /M11 – DC11_matc /M11 – DC11_matc /M6 – OF6_match /M6 – PR6_match	vn n Trigger Sele ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾	-n/n = Value at	-	Resets
' = Bit is set it 7-6 it 5-0	Unimplemen TRIGSEL<5: 101101 = PW 10100 = PW 10101 = PW 101001 = PW 101000 = PW 100101 = PW 100111 = PW 100101 = PW 100101 = PW 100011 = PW 100011 = PW 100011 = PW	'0' = Bit is cleare ted: Read as '0' D:: Auto-Conversio /M12 – OF12_mato /M12 – PH12_mato /M12 – DC12_mato /M12 – DC12_mato /M11 – OF11_mato /M11 – PH11_mato /M11 – DC11_mato /M11 – DC11_mato /M10 – OF6_match /M6 – PH6_match /M6 – PR6_match	n Trigger Sele _{Sh} (2) _{Sh} (2) _{Sh} (2) _{Sh} (2) _{Sh} (2) _{Sh} (2) _{Sh} (2) _{Sh}			Resets
it 7-6 it 5-0	TRIGSEL<5:0 101101 = PW 101100 = PW 101011 = PW 101001 = PW 101000 = PW 100101 = PW 100111 = PW 100101 = PW 100101 = PW 100011 = PW 100010 = PW 100001 = PW	ted: Read as '0' D>: Auto-Conversio /M12 – OF12_mato /M12 – PH12_mato /M12 – DC12_mato /M11 – OF11_mato /M11 – PH11_mato /M11 – PR11_mato /M11 – DC11_mato /M11 – DC11_mato /M6 – PH6_match /M6 – PR6_match	n Trigger Sele ₂ h(2) ₂ h(2) ₂ h(2) ₂ h(2) ₂ h(2) ¹ ¹ ¹ ¹	ction bits ⁽¹⁾		
it 5-0	TRIGSEL<5:0 101101 = PW 101100 = PW 101011 = PW 101001 = PW 101000 = PW 100101 = PW 100111 = PW 100101 = PW 100101 = PW 100011 = PW 100010 = PW 100001 = PW	D>: Auto-Conversio /M12 – OF12_mato /M12 – PH12_mato /M12 – PR12_mato /M12 – DC12_mato /M11 – OF11_mato /M11 – PH11_mato /M11 – PR11_mato /M11 – DC11_mato /M6 – OF6_match /M6 – PR6_match	ch(2) ch(2) ch(2) ch(2) ch(2) ch ch ch ch	ction bits ⁽¹⁾		
	101101 = PW 101100 = PW 101011 = PW 101001 = PW 101000 = PW 100101 = PW 100111 = PW 100101 = PW 100100 = PW 100011 = PW 100010 = PW	VM12 – OF12_mato VM12 – PH12_mato VM12 – PR12_mato VM12 – DC12_mato VM11 – OF11_mato VM11 – PH11_mato VM11 – PR11_mato VM11 – DC11_mato VM6 – OF6_match VM6 – PR6_match	ch(2) ch(2) ch(2) ch(2) ch(2) ch ch ch ch	ction bits ⁽¹⁾		
	101100 = PW 101011 = PW 101001 = PW 101000 = PW 100101 = PW 100111 = PW 100101 = PW 100100 = PW 100011 = PW 100011 = PW 100010 = PW	VM12 – PH12_mato VM12 – PR12_mato VM12 – DC12_mato VM11 – OF11_mato VM11 – PH11_mato VM11 – PR11_mato VM11 – DC11_mato VM6 – OF6_match VM6 – PH6_match	ch ⁽²⁾ ch ⁽²⁾ ch ⁽²⁾ ch ch ch			
	101011 = PW 101001 = PW 101001 = PW 100101 = PW 100111 = PW 100101 = PW 100101 = PW 100011 = PW 100011 = PW 100010 = PW	VM12 – PR12_mato VM12 – DC12_mato VM11 – OF11_mato VM11 – PH11_mato VM11 – PR11_mato VM11 – DC11_mato VM6 – OF6_match VM6 – PH6_match VM6 – PR6_match	_{ch} (2) _{Ch} (2) :h :h			
	101010 = PW 101001 = PW 101000 = PW 100111 = PW 100101 = PW 100101 = PW 100011 = PW 100011 = PW 100010 = PW	VM12 – DC12_mato VM11 – OF11_mato VM11 – PH11_mato VM11 – PR11_mato VM11 – DC11_mato VM6 – OF6_match VM6 – PH6_match VM6 – PR6_match	_{ch} (2) :h :h :h			
	101001 = PW 101000 = PW 100111 = PW 100101 = PW 100101 = PW 100100 = PW 100011 = PW 100010 = PW	/M11 – OF11_matc /M11 – PH11_matc /M11 – PR11_matc /M11 – DC11_matc /M6 – OF6_match /M6 – PH6_match /M6 – PR6_match	:h :h :h			
	101000 = PW 100111 = PW 100101 = PW 100101 = PW 100100 = PW 100011 = PW 100010 = PW	/M11 – PH11_matc /M11 – PR11_matc /M11 – DC11_matc /M6 – OF6_match /M6 – PH6_match /M6 – PR6_match	:h :h			
	100111 = PW 100110 = PW 100101 = PW 100100 = PW 100011 = PW 100010 = PW 100001 = PW	/M11 – PR11_matc /M11 – DC11_matc /M6 – OF6_match /M6 – PH6_match /M6 – PR6_match	h			
	100110 = PW 100101 = PW 100100 = PW 100011 = PW 100010 = PW 100001 = PW	/M11 – DC11_matc /M6 – OF6_match /M6 – PH6_match /M6 – PR6_match				
	100101 = PW 100100 = PW 100011 = PW 100010 = PW 100001 = PW	/M6 – OF6_match /M6 – PH6_match /M6 – PR6_match				
	100011 = PV 100010 = PV 100001 = PV	/M6 – PR6_match				
	100010 = PV 100001 = PV	—				
	100001 = PW	M6 – DC6 match				
		—				
		/M5 – PH5_match				
		M5 – PH5_match				
		/M5 – PH5_match /M5 – PH5 match				
		/M10 – PWM10OU	(2)			
		M9 – PWM9OUT	•			
		M4 – PWM4OUT				
	011010 = PW	M3 – PWM3OUT				
	011001 = CC	P8 – CCP8_trigger	₍ 2)			
		P7 – CCP7_trigger				
		P2 – CCP2_trigger				
		P1 – CCP1_trigger	ſ			
		C4 – LC4_out				
		C3 – LC3_out C2 – LC2_out				
		$C_2 = LC_2_0ut$ $C_1 = LC_1_out$				
		mparator C8 – synd	C80UT ⁽²⁾			
	010000 = Co	mparator C7 – synd	C70UT ⁽²⁾			
		mparator C6 – synd	-			
	001110 = Co	mparator C5 - synd	C5OUT			
		mparator C4 – synd				
		mparator C3 – synd				
		mparator C2 – synd	_			
		mparator C1 – syno ner8 – T8_postscalo				
		ner6 – T6_postscal				
		ner5 – T5 overflow				
		ner4 – T4_postscal				
		ner3 – T3_overflow				
	000100 = Tin	ner2 – T2_postscal	ed			
		ner1 – T1_overflow				
		ner0 – T0_overflow				
		CACT – ADCACTF				
	000000 = No	Auto-conversion T	ngger selected	L		

Note1:This is a rising edge sensitive input for all sources.2:PIC16(L)F1777/9 only.

REGISTER 16-4: ADRESH: ADC RESULT REGISTER HIGH (ADRESH) ADFM = 0

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
			ADRE	S<9:2>			
bit 7							bit 0
Legend:							
R = Readable	Readable bit W = Writable bit U = Unimplemented bit, read as '0'						
u = Bit is unch	anged	x = Bit is unkn	nknown -n/n = Value at POR and BOR/Value at all othe			other Resets	
'1' = Bit is set		'0' = Bit is clea	ared				

bit 7-0 **ADRES<9:2>**: ADC Result Register bits Upper eight bits of 10-bit conversion result

REGISTER 16-5: ADRESL: ADC RESULT REGISTER LOW (ADRESL) ADFM = 0

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
ADRES<1:0>		—	—	—	—	—	—
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-6 ADRES<1:0>: ADC Result Register bits Lower two bits of 10-bit conversion result

bit 5-0 **Reserved**: Do not use.

REGISTER 16-6: ADRESH: ADC RESULT REGISTER HIGH (ADRESH) ADFM = 1

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	
—				—	— — ADRES<9:8>			
bit 7				•			bit 0	
Legend:								
R = Readable bit W = Writable bit			bit	U = Unimplemented bit, read as '0'				
u = Bit is unchanged x = Bit is unknown		nown	-n/n = Value at POR and BOR/Value at all other Resets					
'1' = Bit is set '0		'0' = Bit is clea	ared					
-								

bit 7-2 Reserved: Do not use. bit 1-0 ADRES<9:8>: ADC Result Register bits Upper two bits of 10-bit conversion result

REGISTER 16-7: ADRESL: ADC RESULT REGISTER LOW (ADRESL) ADFM = 1

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	
ADRES<7:0>								
bit 7 b								

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 ADRES<7:0>: ADC Result Register bits Lower eight bits of 10-bit conversion result

16.4 ADC Acquisition Requirements

For the ADC to meet its specified accuracy, the charge holding capacitor (CHOLD) must be allowed to fully charge to the input channel voltage level. The Analog Input model is shown in Figure 16-4. The source impedance (Rs) and the internal sampling switch (Rss) impedance directly affect the time required to charge the capacitor CHOLD. The sampling switch (Rss) impedance varies over the device voltage (VDD), refer to Figure 16-4. The maximum recommended impedance for analog sources is 10 k Ω . As the

source impedance is decreased, the acquisition time may be decreased. After the analog input channel is selected (or changed), an ADC acquisition must be done before the conversion can be started. To calculate the minimum acquisition time, Equation 16-1 may be used. This equation assumes that 1/2 LSb error is used (1,024 steps for the ADC). The 1/2 LSb error is the maximum error allowed for the ADC to meet its specified resolution.

EQUATION 16-1: ACQUISITION TIME EXAMPLE^(1,2,3)

Assumptions: Temperature =
$$50^{\circ}C$$
 and external impedance of $10k\Omega 5.0V$ VDD
 $TACQ = Amplifier Settling Time + Hold Capacitor Charging Time + Temperature Coefficient$
 $= TAMP + TC + TCOFF$
 $= 2\mu s + TC + [(Temperature - 25^{\circ}C)(0.05\mu s/^{\circ}C)]$

The value for TC can be approximated with the following equations:

$$V_{APPLIED}\left(1 - \frac{1}{(2^{n+1}) - 1}\right) = V_{CHOLD} \qquad ;[1] V_{CHOLD} charged to within 1/2 lsb$$

$$V_{APPLIED}\left(1 - e^{\frac{-Tc}{RC}}\right) = V_{CHOLD} \qquad ;[2] V_{CHOLD} charge response to V_{APPLIED} V_{APPLIED}\left(1 - \frac{1}{(2^{n+1}) - 1}\right) \qquad ;combining [1] and [2]$$

Note: Where n = number of bits of the ADC.

Solving for TC:

$$Tc = -CHOLD(RIC + RSS + RS) \ln(1/2047)$$

= $-10pF(1k\Omega + 7k\Omega + 10k\Omega) \ln(0.0004885)$
= $1.37\mu s$

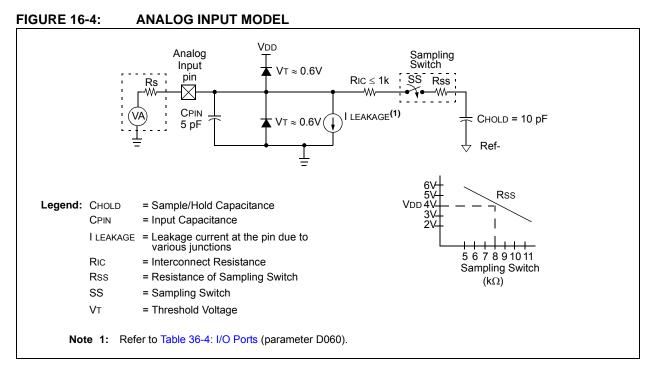
Therefore:

$$TACQ = 2\mu s + 892ns + [(50^{\circ}C - 25^{\circ}C)(0.05\mu s/^{\circ}C)]$$

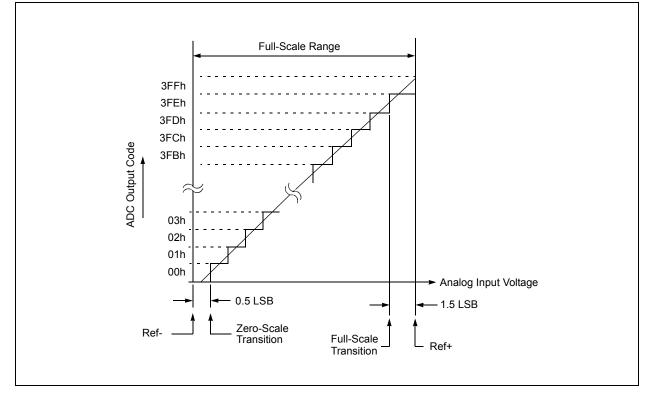
= 4.62\mu s

Note 1: The reference voltage (VREF) has no effect on the equation, since it cancels itself out.

- 2: The charge holding capacitor (CHOLD) is not discharged after each conversion.
- **3:** The maximum recommended impedance for analog sources is $10 \text{ k}\Omega$. This is required to meet the pin leakage specification.







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Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ADCON0			CHS	<5:0>			GO/DONE	ADON	232
ADCON1	ADFM		ADCS<2:0>		—	ADNREF	ADPRE	F<1:0>	234
ADCON2	_	_			TRIGS	EL<5:0>			235
ADRESH	ADC Result	ADC Result Register High							
ADRESL	ADC Result	Register Low	v						236, 237
ANSELA	—	_	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0	177
ANSELB	—	_	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	182
ANSELC	ANSC7	ANSC6	ANSC5	ANSC4	ANSC3	ANSC2	—	_	187
FVRCON	FVREN	FVRRDY	TSEN	TSRNG	CDAFVR<1:0> ADFVR<1:0>			R<1:0>	223
DAC1CON0	EN	FM	OE1	OE2	PSS	<1:0>	NSS	<1:0>	244
DAC2CON0	EN	FM	OE1	OE2	PSS	<1:0>	NSS	<1:0>	244
DAC5CON0	EN	FM	OE1	OE2	PSS	<1:0>	NSS	<1:0>	244
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	132
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	CCP1IE	TMR2IE	TMR1IE	133
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	CCP1IF	TMR2IF	TMR1IF	139
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	176
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB6	TRISB6	TRISB1	TRISB0	181
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	186

TABLE 16-3: SUMMARY OF REGISTERS ASSOCIATED WITH ADC

Legend: x = unknown, u = unchanged, - = unimplemented read as '0', q = value depends on condition. Shaded cells are not used for the ADC module.

17.0 5-BIT DIGITAL-TO-ANALOG CONVERTER (DAC) MODULE

The Digital-to-Analog Converter supplies a variable voltage reference, ratiometric with the input source, with 32 selectable output levels.

The input of the DAC can be connected to:

- External VREF pins
- VDD supply voltage
- FVR (Fixed Voltage Reference)

The output of the DAC can be configured to supply a reference voltage to the following:

- Comparator positive input
- Operational amplifier inverting and non-inverting inputs
- ADC input channel
- DACxOUTx pin

TABLE 17-1: AVAILABLE 5-BIT DACS

Device	DAC3	DAC4	DAC7	DAC8
PIC16(L)F1778	•	•	•	
PIC16(L)F1777/9	•	•	•	•

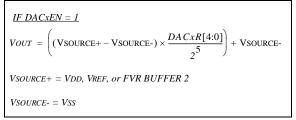
The Digital-to-Analog Converter (DAC) is enabled by setting the EN bit of the DACxCON0 register.

17.1 Output Voltage Selection

The DAC has 32 voltage level ranges. The 32 levels are set with the REF<4:0> bits of the DACxREF register.

The DAC output voltage is determined by Equation 17-1:

EQUATION 17-1: DAC OUTPUT VOLTAGE



17.2 Ratiometric Output Level

The DAC output value is derived using a resistor ladder with each end of the ladder tied to a positive and negative voltage reference input source. If the voltage of either input source fluctuates, a similar fluctuation will result in the DAC output value.

The value of the individual resistors within the ladder can be found in Table 36-20: 10-bit Digital-to-Analog Converter (DAC) Specifications.

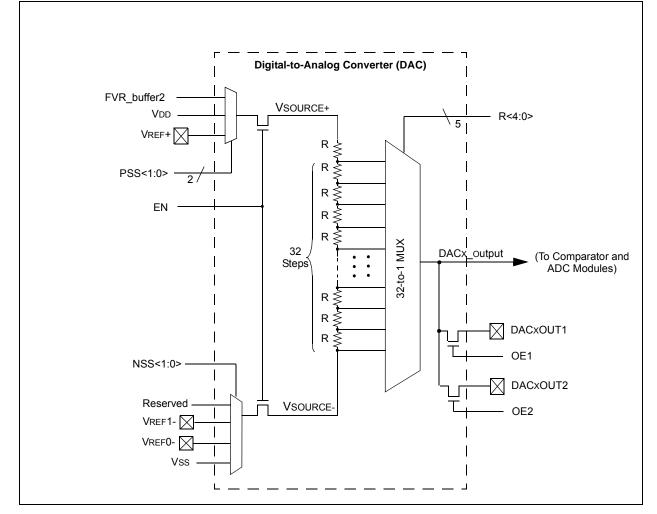
17.3 DAC Voltage Reference Output

The DAC voltage can be output to the DACxOUTx pin by setting the OEx bit of the DACxCON0 register. Selecting the DAC voltage for output on the DACxOUTx pin automatically overrides the digital output buffer and digital input threshold detector functions of that pin. Reading the DACxOUTx pin when it has been configured for DAC voltage output will always return a '0'.

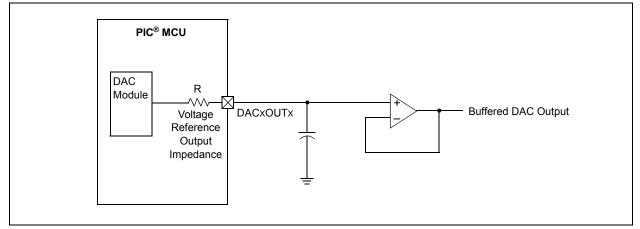
Due to the limited current drive capability, a buffer must be used on the DAC voltage output for external connections to the DACxOUTx pin. Figure 17-2 shows an example buffering technique.

PIC16(L)F1777/8/9

FIGURE 17-1: DIGITAL-TO-ANALOG CONVERTER BLOCK DIAGRAM







17.4 Operation During Sleep

The DAC continues to function during Sleep. When the device wakes up from Sleep through an interrupt or a Watchdog Timer time-out, the contents of the DACxCON0 register are not affected. To minimize current consumption in Sleep mode, the voltage reference should be disabled.

17.5 Effects of a Reset

A device Reset affects the following:

- DAC is disabled.
- DAC output voltage is removed from the DACxOUTx pin.
- The REF<4:0> voltage reference control bits are cleared.

17.6 Register Definitions: DAC Control

Long bit name prefixes for the 5-bit DAC peripherals are shown in Table 17-2. Refer to **Section 1.1 "Register and Bit naming conventions"** for more information

TABLE 17-2:

Peripheral	Bit Name Prefix
DAC3	DAC3
DAC4	DAC4
DAC7	DAC7
DAC8 ⁽¹⁾	DAC8

Note 1: PIC16(L)F1777/9 only.

REGISTER 17-1: DACxCON0: DACx CONTROL REGISTER 0

R/W-0/0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	U-0	R/W-0/0
EN	—	OE1	OE2	PSS<1:0>		NSS	S<1:0>
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7	EN: DAC Enable bit 1 = DAC is enabled 0 = DAC is disabled
bit 6	Unimplemented: Read as '0'
bit 5	 OE1: DAC Voltage Output Enable bit 1 = DAC voltage level is also an output on the DACxOUT1 pin 0 = DAC voltage level is disconnected from the DACxOUT1 pin
bit 4	 OE2: DAC Voltage Output Enable bit 1 = DAC voltage level is also an output on the DACxOUT2 pin 0 = DAC voltage level is disconnected from the DACxOUT2 pin
bit 3-2	<pre>PSS<1:0>: DAC Positive Source Select bits 11 = Reserved, do not use 10 = FVR Buffer2 output 01 = VREF+ pin 00 = VDD</pre>
bit 1-0	NSS<1:0>: DAC Negative Source Select bits 11 = Reserved, do not use 10 = DACxREF1- (DAC7/8) or Reserved (DAC3/4) 01 = DACxREF0- 00 = AGND (AVSS)

REGISTER 17-2: DACxREF: DACx REFERENCE VOLTAGE OUTPUT SELECT REGISTER

U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	—	—			REF<4:0>		
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-5 Unimplemented: Read as '0'

bit 4-0 **REF<4:0>:** DACx Reference Voltage Output Select bits (See Equation 17-1)

TABLE 17-3: SUMMARY OF REGISTERS ASSOCIATED WITH THE DACX MODULE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on page
DAC3CON0	EN	_	OE1	OE2	PSS	<1:0>	NSS	<1:0>	244
DAC4CON0	EN		OE1	OE2	PSS<1:0>		NSS<1:0>		244
DAC7CON0	EN		OE1	OE2	PSS<1:0>		NSS<1:0>		244
DAC8CON0 ⁽¹⁾	EN		OE1	OE2	PSS<1:0>		NSS<1:0>		244
DAC3REF						245			
DAC4REF						245			
DAC7REF				REF<4:0>					245
DAC8REF ⁽¹⁾						REF<4:0>			245

Legend: — = Unimplemented location, read as '0'. Shaded cells are not used with the DACx module.

Note 1: PIC16LF1777/9 only.

18.0 10-BIT DIGITAL-TO-ANALOG CONVERTER (DAC) MODULE

The 10-bit Digital-to-Analog Converter (DAC) supplies a variable voltage reference, ratiometric with the input source, with 1024 selectable output levels.

The input of the DAC can be connected to:

- · External VREF pins
- VDD supply voltage
- FVR (Fixed Voltage Reference)

The output of the DAC can be configured to supply a reference voltage to the following:

- Comparator positive input
- · ADC input channel
- DACxOUT1 pin
- Op Amp

The Digital-to-Analog Converter is enabled by setting the EN bit of the DACxCON0 register.

TABLE 18-1: AVAILABLE 10-BIT DACS

Device	DAC1	DAC2	DAC5	DAC6
PIC16(L)F1778	•	•	•	
PIC16(L)F1777/9	•	•	•	•

18.1 Output Voltage Level Selection

The DAC has 1024 voltage levels that are set by the 10-bit reference selection word contained in the DACxREFH and DACxREFL registers. This 10-bit word can be left or right justified. See Section 18.4 "DAC Reference Selection Justification" for more detail.

The DAC output voltage can be determined with Equation 18-1.

18.2 Ratiometric Output Voltage

The DAC output voltage is derived using a resistor ladder with each end of the ladder tied to a positive and negative voltage source. If the voltage of either input source fluctuates, a similar fluctuation will result in the DAC output value.

The value of the individual resistors within the ladder can be found in Table 36-20.

18.3 DAC Output

The DAC voltage is always available to the internal peripherals that use it. The DAC voltage can be output to the DACxOUTx pin by setting the OEx bit of the DACxCON0 register. Selecting the DAC voltage for output on the DACxOUTx pin automatically overrides the digital output buffer and digital input threshold detector functions of that pin. Reading the DACxOUTx pin when it has been configured for DAC voltage output will always return a '0'.

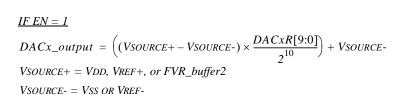
Due to the limited current drive capability, a buffer must be used on the DAC voltage output for external connections to either DACxOUTx pin. Figure 18-3 shows a buffering technique example.

18.4 DAC Reference Selection Justification

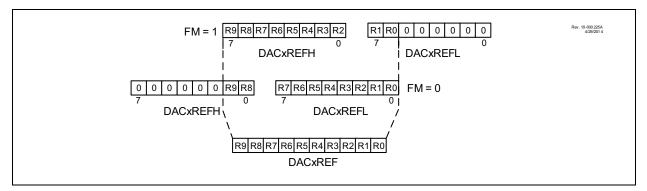
The DAC reference selection can be configured to be left or right justified. When the FM bit of the DACxCON0 register is set, the 10-bit word is left justified such that the eight Most Significant bits fill the DACxREFH register and the two Least Significant bits are left justified in the DACxREFL register. When the FM bit is cleared, the 10-bit word is right justified such that the eight Least Significant bits fill the DACxREFL register and the two Most Significant bits are right justified in the DACxREFH register. Refer to Figure 18-1.

The DACxREFL and DACxREFH registers are double buffered. Writing to either register does not take effect immediately. Writing a '1' to the DACxLD bit of the DACLD register transfers the contents of the DACxREFH and DACxREFL registers to the buffers, thereby changing all 10-bits of the DAC reference selection simultaneously.

EQUATION 18-1: DAC OUTPUT VOLTAGE









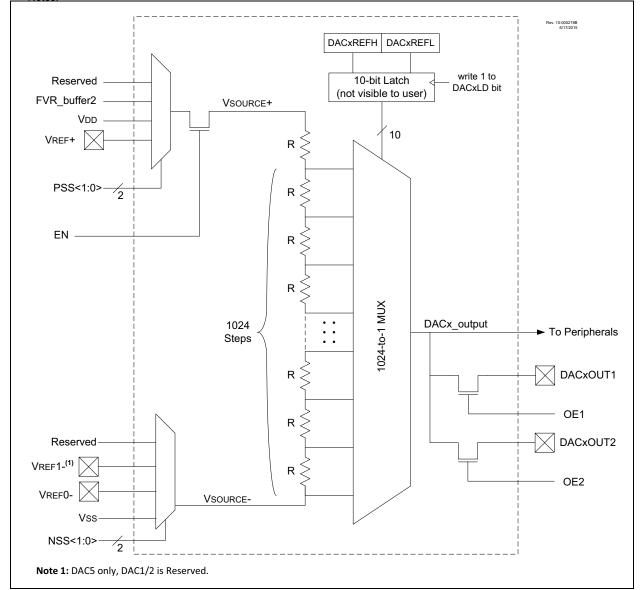
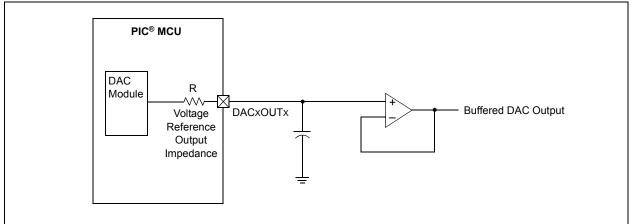


FIGURE 18-3: VOLTAGE REFERENCE OUTPUT BUFFER EXAMPLE



18.5 Operation During Sleep

When the device wakes up from Sleep as the result of an interrupt or a Watchdog Timer time-out, the contents of the DACxCON0 register are not affected. To minimize current consumption in Sleep mode, the voltage reference should be disabled.

18.6 Effects of a Reset

A device Reset affects the following:

- · DAC is disabled
- DAC output voltage is removed from the DACxOUTx pin
- The REF<9:0> reference selection bits are cleared

18.7 Register Definitions: DAC Control

Long bit name prefixes for the 10-bit DAC peripherals are shown in Table 18-2. Refer to **Section 1.1** "**Register and Bit naming conventions**" for more information

TABLE 18-2:

Peripheral	Bit Name Prefix
DAC1	DAC1
DAC2	DAC2
DAC5	DAC5
DAC6 ⁽¹⁾	DAC6

Note 1: PIC16(L)F1777/9 only.

REGISTER 18-1: DACxCON0: DAC CONTROL REGISTER 0

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
EN	FM	OE1	OE2	PSS<1:0>		NSS	<1:0>
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7	EN: DAC Enable bit
	1 = DACx is enabled
	0 = DACx is disabled
bit 6	FM: DAC Reference Format bit
	1 = DACx reference selection is left justified
	0 = DACx reference selection is right justified
bit 5	OE1: DAC Voltage Output Enable bit
	1 = DACx voltage level is also an output on the DACxOUT1 pin
	0 = DACx voltage level is disconnected from the DACxOUT1 pin
bit 4	OE2: DAC Voltage Output Enable bit
	1 = DACx voltage level is also an output on the DACxOUT2 pin
	0 = DACx voltage level is disconnected from the DACxOUT2 pin
bit 3-2	PSS<1:0>: DAC Positive Source Select bits
	11 = DACxREF1+ (DAC5/6) or Reserved (DAC1/2)
	$10 = FVR_buffer2$
	01 = DACxREF0+
	00 = VDD
bit 1-0	NSS<1:0>: DAC Negative Source Select bit
	11 = Reserved. Do not use.
	10 = DACxREF1- (DAC5/6) or Reserved (DAC1/2)
	01 = DACxREF0-
	00 = Agnd(AVss)

REGISTER 18-2: DACxREFH: DAC REFERENCE VOLTAGE SELECT HIGH REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0			
		RE	EF<9:x> (x De	pends on FM b	it)					
bit 7							bit 0			
Legend:										
R = Readable	bit	W = Writable	bit	U = Unimpler	nented bit, read	d as '0'				
u = Bit is unchanged x = Bit is unknown				-n/n = Value at POR and BOR/Value at all other Resets						
'1' = Bit is set		'0' = Bit is clea	ared							
When $FM = 1$	(left justified)									
bit 7-0		DAC Reference = f(REF<9:0>) (•							
When FM = 0	(right justified))								
bit 7-2	2 Unimplemented: Read as '0'									
bit 1-0	·									

REGISTER 18-3: DACxREFL: DAC REFERENCE VOLTAGE SELECT LOW REGISTER

R/W-0/0	R/W-0/0 R/W-0/0 R/W-0/0 F			R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0			
	REF <x-1:0> (x Depends on FM bit)</x-1:0>									
bit 7	bit 7 bit 0									

Legend:						
R = Readable	bit	W = Writable bit	U = Unimplemented bit, read as '0'			
u = Bit is unch	anged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets			
'1' = Bit is set		'0' = Bit is cleared				
When $FM = 1$	(left justified)					
bit 7-6		DAC Reference Voltage Ou = f(REF<9:0>) (See Equa	•			
bit 5-0	Unimpleme	nted: Read as '0'				
<u>When FM = 0 (right justified)</u>						
bit 7-0	REF<7:0>: DAC Reference Voltage Output Select bits DACxOUT1 = f(REF<9:0>) (See Equation 18-1)					

U-0	U-0	U-0	R/W-0/0	U-0	U-0	R/W-0/0	R/W-0/0	
	—	DAC6LD ⁽¹⁾	DAC5LD	—	—	DAC2LD	DAC1LD	
bit 7						bit 0		
Legend:								
R = Readable b	bit	W = Writable	bit	U = Unimpler	mented bit, read	l as '0'		
u = Bit is uncha	anged	x = Bit is unkn	iown	-n/n = Value a	at POR and BO	R/Value at all c	other Resets	
'1' = Bit is set		'0' = Bit is clea	ared	q = value dep	pends on config	uration bits		
bit 7-6	-	ted: Read as 'o						
bit 5		C6 Double Buf						
	1 = DAC6RE by hardw		L values are	transfered to t	he double buffe	r. Bit is cleared	l automatically	
	0 = DAC6RE	FHL:DAC6REF	L double buf	fers remain un	changed.			
bit 4	DAC5LD: DA	C5 Double Buf	fer Load bit					
	1 = DAC5RE by hardw		L values are	transfered to t	he double buffe	r. Bit is cleared	lautomatically	
	0 = DAC5RE	FHL:DAC5REF	L double buf	fers remain un	changed.			
bit 3-2	Unimplemen	ted: Read as ')'					
bit 1	DAC2LD: DA	C2 Double Buf	fer Load bit					
	1 = DAC2RE by hardw		L values are	transfered to t	he double buffe	r. Bit is cleared	lautomatically	
	0 = DAC2RE	FHL:DAC2REF	L double buf	fers remain un	changed.			
bit 0	DAC1LD: DA	C1 Double Buf	fer Load bit					
	1 = DAC1REFHL:DAC1REFL values are transferred to the double buffer. Bit is cleared automatically by hardware.							
		FHL:DAC1RE	L double buf	fers remain un	changed.			
Note 1: PIC	16LF1777/9 or	nly.						

REGISTER 18-4: DACLD: DAC BUFFER LOAD REGISTER

TABLE 18-3: SUMMARY O	F REGISTERS ASSOCIATED WITH THE DACX MODULE
-----------------------	---

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
DAC1CON0	EN	FM	OE1	OE2	PSS	<1:0>	NSS	<1:0>	249
DAC2CON0	EN	FM	OE1	OE2	PSS	<1:0>	NSS	<1:0>	249
DAC5CON0	EN	FM	OE1	OE2	PSS	<1:0>	NSS	<1:0>	249
DAC6CON0 ⁽¹⁾	EN	FM	OE1	OE2	PSS	249			
DAC1REFH	REF<9:x> (x Depends on FM bit)								250
DAC2REFH	REF<9:x> (x Depends on FM bit)								
DAC5REFH	REF<9:x> (x Depends on FM bit)								
DAC6REFH ⁽¹⁾			RE	EF<9:x> (x D	epends on F	M bit)			250
DAC1REFL			REF	= <x-1:0> (x [</x-1:0>	Depends on F	FM bit)			250
DAC2REFL			REF	= <x-1:0> (x [</x-1:0>	Depends on F	=M bit)			250
DAC5REFL	REF <x-1:0> (x Depends on FM bit)</x-1:0>								250
DAC6REFL ⁽¹⁾	REF <x-1:0> (x Depends on FM bit)</x-1:0>								250
DACLD			_	DAC5LD	_		DAC2LD	DAC1LD	251

Legend: — = Unimplemented location, read as '0'. Shaded cells are not used with the DACx module.

Note 1: PIC16LF1777/9 only.

19.0 COMPARATOR MODULE

Comparators are used to interface analog circuits to a digital circuit by comparing two analog voltages and providing a digital indication of their relative magnitudes. Comparators are very useful mixed signal building blocks because they provide analog functionality independent of program execution. The analog comparator module includes the following features:

- Independent comparator control
- Programmable input selection
- · Comparator output is available internally/externally
- Programmable output polarity
- Interrupt-on-change
- Wake-up from Sleep
- Programmable Speed/Power optimization
- · PWM shutdown
- · Programmable and Fixed Voltage Reference

19.1 Comparator Overview

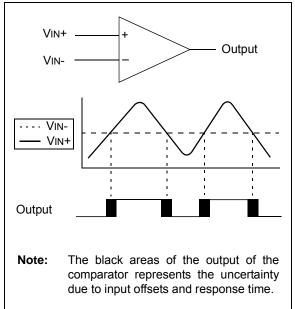
A single comparator is shown in Figure 19-1 along with the relationship between the analog input levels and the digital output. When the analog voltage at VIN+ is less than the analog voltage at VIN-, the output of the comparator is a digital low level. When the analog voltage at VIN+ is greater than the analog voltage at VIN-, the output of the comparator is a digital high level.

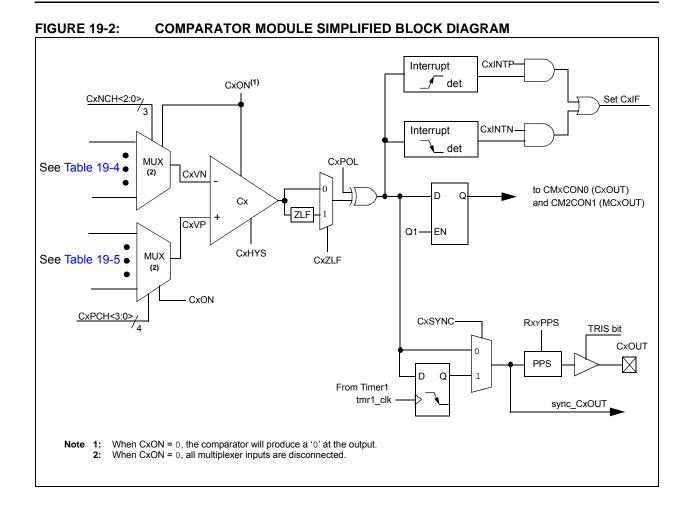
The comparators available for this device are located in Table 19-1.

TABLE 19-1:	AVAILABLE COMPARATORS
-------------	-----------------------

Device	C1	C2	C3	C4	C5	C6	C7	C 8
PIC16(L)F1778	٠	٠	٠	٠	٠	٠		
PIC16(L)F1777/9	٠	٠	٠	٠	٠	٠	٠	•

FIGURE 19-1: SINGLE COMPARATOR





19.2 Comparator Control

Each comparator has two control registers: CMxCON0 and CMxCON1.

The CMxCON0 register (see Register 19-1) contains Control and Status bits for the following:

- Enable
- Output
- Output polarity
- · Zero latency filter
- Speed/Power selection
- Hysteresis enable
- · Output synchronization

The CMxCON1 register (see Register 19-2) contains Control bits for the following:

- · Interrupt enable
- · Interrupt edge polarity
- Positive input channel selection
- Negative input channel selection

19.2.1 COMPARATOR ENABLE

Setting the ON bit of the CMxCON0 register enables the comparator for operation. Clearing the ON bit disables the comparator resulting in minimum current consumption.

19.2.2 COMPARATOR OUTPUT SELECTION

The output of the comparator can be monitored by reading either the OUT bit of the CMxCON0 register or the MCxOUT bit of the CMOUT register. In order to make the output available for an external connection, the following conditions must be true:

- · Desired pin PPS control
- Corresponding TRIS bit must be cleared
- ON bit of the CMxCON0 register must be set

Note 1: The internal output of the comparator is latched with each instruction cycle. Unless otherwise specified, external outputs are not latched.

19.2.3 COMPARATOR OUTPUT POLARITY

Inverting the output of the comparator is functionally equivalent to swapping the comparator inputs. The polarity of the comparator output can be inverted by setting the POL bit of the CMxCON0 register. Clearing the POL bit results in a non-inverted output.

Table 19-2 shows the output state versus input conditions, including polarity control.

TABLE 19-2: COMPARATOR OUTPUT STATE VS. INPUT CONDITIONS

Input Condition	CxPOL	CxOUT
CxVN > CxVP	0	0
CxVN < CxVP	0	1
CxVN > CxVP	1	1
CxVN < CxVP	1	0

19.3 Comparator Hysteresis

A selectable amount of separation voltage can be added to the input pins of each comparator to provide a hysteresis function to the overall operation. Hysteresis is enabled by setting the HYS bit of the CMxCON0 register.

See Comparator Specifications in Table 36-19: Comparator Specifications for more information.

19.4 Timer1 Gate Operation

The output resulting from a comparator operation can be used as a source for gate control of Timer1. See **Section 22.6 "Timer1 Gate"** for more information. This feature is useful for timing the duration or interval of an analog event.

It is recommended that the comparator output be synchronized to Timer1. This ensures that Timer1 does not increment while a change in the comparator is occurring.

19.4.1 COMPARATOR OUTPUT SYNCHRONIZATION

The output from a comparator can be synchronized with Timer1 by setting the SYNC bit of the CMxCON0 register.

Once enabled, the comparator output is latched on the falling edge of the Timer1 source clock. If a prescaler is used with Timer1, the comparator output is latched after the prescaling function. To prevent a race condition, the comparator output is latched on the falling edge of the Timer1 clock source and Timer1 increments on the rising edge of its clock source. See the Comparator Block Diagram (Figure 19-2) and the Timer1 Block Diagram (Figure 22-1) for more information.

19.5 Comparator Interrupt

An interrupt can be generated upon a change in the output value of the comparator for each comparator, a rising edge detector and a falling edge detector are present.

When either edge detector is triggered and its associated enable bit is set (INTP and/or INTN bits of the CMxCON1 register), the Corresponding Interrupt Flag bit (CxIF bit of the PIR2 register) will be set.

To enable the interrupt, you must set the following bits:

- ON and POL bits of the CMxCON0 register
- CxIE bit of the PIE2 register
- INTP bit of the CMxCON1 register (for a rising edge detection)
- INTN bit of the CMxCON1 register (for a falling edge detection)
- PEIE and GIE bits of the INTCON register

The associated interrupt flag bit, CxIF bit of the PIR2 register, must be cleared in software. If another edge is detected while this flag is being cleared, the flag will still be set at the end of the sequence.

Note: Although a comparator is disabled, an interrupt can be generated by changing the output polarity with the POL bit of the CMxCON0 register, or by switching the comparator on or off with the ON bit of the CMxCON0 register.

19.6 Comparator Positive Input Selection

Configuring the PCH<3:0> bits of the CMxPSEL register directs an internal voltage reference or an analog pin to the non-inverting input of the comparator:

- · CxIN+ analog pin
- Programmable ramp generator output
- DAC output
- FVR (Fixed Voltage Reference)
- Vss (Ground)

See **Section 14.0 "Fixed Voltage Reference (FVR)"** for more information on the Fixed Voltage Reference module.

See Section 17.0 "5-Bit Digital-to-Analog Converter (DAC) Module" for more information on the DAC input signal.

Any time the comparator is disabled (CxON = 0), all comparator inputs are disabled.

19.7 Comparator Negative Input Selection

The NCH<3:0> bits of the CMxNSEL register direct an analog input pin and internal reference voltage or analog ground to the inverting input of the comparator:

- CxIN- pin
- FVR (Fixed Voltage Reference)
- Analog Ground

Some inverting input selections share a pin with the operational amplifier output function. Enabling both functions at the same time will direct the operational amplifier output to the comparator inverting input.

Note: To use CxINy+ and CxINy- pins as analog input, the appropriate bits must be set in the ANSEL register and the corresponding TRIS bits must also be set to disable the output drivers.

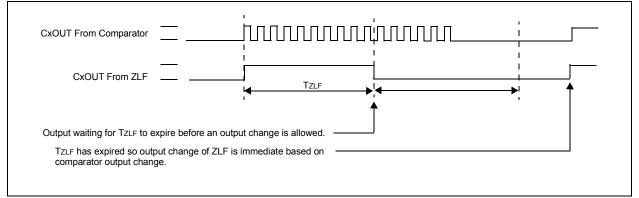
19.8 Comparator Response Time

The comparator output is indeterminate for a period of time after the change of an input source or the selection of a new reference voltage. This period is referred to as the response time. The response time of the comparator differs from the settling time of the voltage reference. Therefore, both of these times must be considered when determining the total response time to a comparator input change. See the Comparator and Voltage Reference Specifications in Table 36-19: Comparator Specifications for more details.

19.9 Zero Latency Filter

In high-speed operation, and under proper circuit conditions, it is possible for the comparator output to oscillate. This oscillation can have adverse effects on the hardware and software relying on this signal. Therefore, a digital filter has been added to the comparator output to suppress the comparator output oscillation. Once the comparator output changes, the output is prevented from reversing the change for a nominal time of 20 ns. This allows the comparator output to stabilize without affecting other dependent devices. Refer to Figure 19-3.





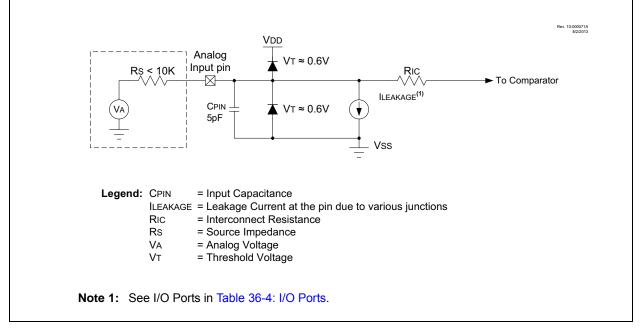
19.10 Analog Input Connection Considerations

A simplified circuit for an analog input is shown in Figure 19-4. Since the analog input pins share their connection with a digital input, they have reverse biased ESD protection diodes to VDD and Vss. The analog input, therefore, must be between Vss and VDD. If the input voltage deviates from this range by more than 0.6V in either direction, one of the diodes is forward biased and a latch-up may occur.

A maximum source impedance of $10 \text{ k}\Omega$ is recommended for the analog sources. Also, any external component connected to an analog input pin, such as a capacitor or a Zener diode, should have very little leakage current to minimize inaccuracies introduced.

- Note 1: When reading a PORT register, all pins configured as analog inputs will read as a '0'. Pins configured as digital inputs will convert as an analog input, according to the input specification.
 - 2: Analog levels on any pin defined as a digital input, may cause the input buffer to consume more current than is specified.





19.11 Register Definitions: Comparator Control

Long bit name prefixes for the Comparator peripherals are shown in Table 19-3. Refer to **Section 1.1.2.2 "Long Bit Names"** for more information

TABLE 19-3:

Peripheral	Bit Name Prefix			
Comparator 1	C1			
Comparator 2	C2			
Comparator 3	C3			
Comparator 4	C4			
Comparator 5	C5			
Comparator 6	C6			
Comparator 7 ⁽¹⁾	C7			
Comparator 8 ⁽¹⁾	C8			

Note 1: PIC16(L)F1777/9 only.

REGISTER 19-1: CMxCON0: COMPARATOR Cx CONTROL REGISTER 0

R/W-0/0	R-0/0	U-0	R/W-0/0	R/W-0/0	R/W-1/1	R/W-0/0	R/W-0/0
ON	OUT	—	POL	ZLF	—	HYS	SYNC
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7	ON: Comparator Enable bit 1 = Comparator is enabled 0 = Comparator is disabled and consumes no active power
bit 6	OUT: Comparator Output bit $If POL = 1$ (inverted polarity): $1 = CxVP < CxVN$ $0 = CxVP > CxVN$ $If POL = 0$ (non-inverted polarity): $1 = CxVP > CxVN$ $0 = CxVP > CxVN$ $0 = CxVP < CxVN$
bit 5	Reserved: Read as '1'. Maintain this bit set.
bit 4	 POL: Comparator Output Polarity Select bit 1 = Comparator output is inverted 0 = Comparator output is not inverted
bit 3	ZLF: Comparator Zero Latency Filter Enable bit 1 = Comparator output is filtered 0 = Comparator output is unfiltered
bit 2	Reserved: Read as '1'. Maintain this bit set.
bit 1	 HYS: Comparator Hysteresis Enable bit 1 = Comparator hysteresis enabled 0 = Comparator hysteresis disabled
bit 0	 SYNC: Comparator Output Synchronous Mode bit 1 = Comparator output to Timer1 and I/O pin is synchronous to changes on Timer1 clock source. Output updated on the falling edge of Timer1 clock source. 0 = Comparator output to Timer1 and I/O pin is asynchronous

U-0	U-0	U-0							
		0-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0		
	_	—	_	_	—	INTP	INTN		
bit 7	-						bit 0		
Legend:									
R = Readable bit W = Writable bit				U = Unimplemented bit, read as '0'					
u = Bit is unch	u = Bit is unchanged x = Bit is unknown		iown	-n/n = Value a	at POR and BO	R/Value at all c	other Resets		
'1' = Bit is set		'0' = Bit is clea	ared						
bit 7-2	bit 7-2 Unimplemented: Read as '0'								
bit 1	INTP: Comparator Interrupt on Positive Going Edge Enable bits								

REGISTER 19-2: CMxCON1: COMPARATOR Cx CONTROL REGISTER 1

1 = The CxIF interrupt flag will be set upon a positive going edge of the CxOUT bit

0 = No interrupt flag will be set on a positive going edge of the CxOUT bit

bit 0 INTN: Comparator Interrupt on Negative Going Edge Enable bits

1 = The CxIF interrupt flag will be set upon a negative going edge of the CxOUT bit

0 = No interrupt flag will be set on a negative going edge of the CxOUT bit

REGISTER 19-3: CMxNSEL: COMPARATOR Cx NEGATIVE CHANNEL SELECT REGISTER

U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0		
—	_	—	_		NCH<3:0>				
bit 7					bit (
Legend:									
R = Readable	bit	W = Writable	bit	U = Unimplemented bit, read as '0'					
u = Bit is uncha	anged	x = Bit is unkr	nown	-n/n = Value at POR and BOR/Value at all other Resets					

bit 7-4 Unimplemented: Read as '0'

'1' = Bit is set

bit 3-0 NCH<3:0>: Comparator Negative Input Channel Select bits CxVN connects to input source indicated by Table 19-4: Negative Input Sources

'0' = Bit is cleared

NCH<3:0>	C1, C2, C3, C4	C5, C6, C7 ⁽¹⁾ , C8 ⁽¹⁾
1111	Reserved. Do not use	Reserved. Do not use
1110	Reserved. Do not use	Reserved. Do not use
1101	Reserved. Do not use	Reserved. Do not use
1100	Reserved. Do not use	Reserved. Do not use
1011	Reserved. Do not use	Reserved. Do not use
1010	OPA2IN- pin	OPA4IN- pin ⁽¹⁾
1001	OPA1IN- pin	OPA3IN- pin
1000	AGND	AGND
0111	PRG2_out	PRG4_out ⁽¹⁾
0110	PRG1_out	PRG3_out
0101	FVR_Buffer2	FVR_Buffer2
0100	CxIN4- pin	CxIN4- (OPA4OUT) pin ⁽¹⁾
0011	CxIN3- (OPA2OUT) pin	CxIN3- pin
0010	CxIN2- pin	CxIN2- pin
0001	CxIN1- (OPA1OUT) pin	CxIN1- (OPA3OUT) pin
0000	CxIN0- pin	CxIN0- pin

TABLE 19-4: NEGATIVE INPUT SOURCES

Note 1: PIC16(L)F1777/9 only.

REGISTER 19-4: CMxPSEL: COMPARATOR Cx POSITIVE CHANNEL SELECT REGISTER 1

U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	_	_	—	PCH<3:0>			
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-4 Unimplemented: Read as '0'

bit 3-0 PCH<3:0>: Comparator Positive Input Channel Select bits CxVP connects to input source indicated by Table 19-5: Positive Input Sources

TABLE 19-5: POSITIVE INPUT SOURCES

PCH<3:0>	C1, C2, C3, C4	C5, C6, C7 ⁽¹⁾ , C8 ⁽¹⁾		
1111	Reserved. Do not use	Reserved. Do not use		
1110	Reserved. Do not use	Reserved. Do not use		
1101	Reserved. Do not use	Reserved. Do not use		
1100	Reserved. Do not use	Reserved. Do not use		
1011	Reserved. Do not use	Reserved. Do not use		
1010	Reserved. Do not use	Reserved. Do not use		
1001	AGND	AGND		
1000	DAC4_out	DAC8_out ⁽¹⁾		
0111	DAC3_out	DAC7_out		
0110	DAC2_out	DAC6_out ⁽¹⁾		
0101	DAC1_out	DAC5_out		
0100	PRG2_out	PRG4_out ⁽¹⁾		
0011	PRG1_out	PRG3_out		
0010	FVR_Buffer2	FVR_Buffer2		
0001	CxIN1+ pin	CxIN1+ pin		
0000	CxIN0+ pin	CxIN0+ pin		

Note 1: PIC16(L)F1777/9 only.

Note: There are no long and short bit name variants for the following mirror register

REGISTER 19-5: CMOUT: COMPARATOR OUTPUT REGISTER

R-0/0	R-0/0	R-0/0	R-0/0	R-0/0	R-0/0	R-0/0	R-0/0
MC8OUT ⁽¹⁾	MC7OUT ⁽¹⁾	MC6OUT	MC5OUT	MC4OUT	MC3OUT	MC2OUT	MC1OUT
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7	MC8OUT: Mirror	Copy of C8OUT bit ⁽¹⁾

- bit 6 **MC7OUT:** Mirror Copy of C7OUT bit⁽¹⁾
- bit 5 MC6OUT: Mirror Copy of C6OUT bit
- bit 4 MC5OUT: Mirror Copy of C5OUT bit
- bit 3 MC4OUT: Mirror Copy of C4OUT bit
- bit 2 MC3OUT: Mirror Copy of C3OUT bit
- bit 1 MC2OUT: Mirror Copy of C2OUT bit
- bit 0 MC10UT: Mirror Copy of C10UT bit

Note 1: PIC16LF1777/9 only.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ANSELA	_	_	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0	177
ANSELB	_	_	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	182
ANSELC	ANSC7	ANSC6	ANSC5	ANSC4	ANSC3	ANSC2	_	_	187
CM1CON0	ON	OUT	_	POL	ZLF	Reserved	HYS	SYNC	258
CM2CON0	ON	OUT	_	POL	ZLF	Reserved	HYS	SYNC	258
CM3CON0	ON	OUT	_	POL	ZLF	Reserved	HYS	SYNC	258
CM4CON0	ON	OUT	_	POL	ZLF	Reserved	HYS	SYNC	258
CM5CON0	ON	OUT	_	POL	ZLF	Reserved	HYS	SYNC	258
CM6CON0	ON	OUT		POL	ZLF	Reserved	HYS	SYNC	258
CM1CON1	_	_	_	_	_	_	INTP	INTN	259
CM2CON1	_	_	_	_	_	_	INTP	INTN	259
CM3CON1	_	_	_	_	_	_	INTP	INTN	259
CM4CON1	_	_	_	_		_	INTP	INTN	259
CM5CON1		_		_			INTP	INTN	259
CM6CON1	_	_	_	_		_	INTP	INTN	259
CM7CON1 ⁽¹⁾	_	_	_	_	_	_	INTP	INTN	259
CM8CON1 ⁽¹⁾	_	_	_	_		_	INTP	INTN	259
CM1NSEL	_	_	_	_		NCH<3:0>		260	
CM2NSEL	_	_	_	_	NCH<3:0>				260
CM3NSEL	_	_	_	_	NCH<3:0>				260
CM4NSEL	_	_	_	_	NCH<3:0>				260
CM5NSEL	_	_	_	_	NCH<3:0>				260
CM6NSEL	_	_	_	_	NCH<3:0>				260
CM7NSEL ⁽¹⁾	_	_	_	_	NCH<3:0>				260
CM8NSEL ⁽¹⁾	_	_	_	_	NCH<3:0>				260
CM1PSEL	_	_	_	_	PCH<3:0>				261
CM2PSEL	_	_	_	_	PCH<3:0>				261
CM3PSEL	_	_	_	_	PCH<3:0>				261
CM4PSEL	_	_	_	_		PCH	<3:0>		261
CM5PSEL	_	_	_	_		PCH	<3:0>		261
CM6PSEL	_	_	_	_		PCH	<3:0>		261
CM7PSEL ⁽¹⁾	_	_	_	_		PCH	<3:0>		261
CM8PSEL(1)	_	_	_	_		PCH	<3:0>		261
CMOUT	MC8OUT ⁽¹⁾	MC7OUT ⁽¹⁾	MC6OUT	MC5OUT	MC4OUT	MC3OUT	MC2OUT	MC1OUT	262
FVRCON	FVREN	FVRRDY	TSEN	TSRNG		/R<1:0>		R<1:0>	223
DAC1CON0	EN	FM	OE1	OE2	PSS	<1:0>	NSS	<1:0>	249
DAC2CON0	EN	FM	OE1	OE2	PSS	<1:0>	NSS	<1:0>	249
DAC5CON0	EN	FM	OE1	OE2	PSS	<1:0>	NSS	<1:0>	249
DAC6CON0 ⁽¹⁾	EN	FM	OE1	OE2		PSS<1:0> PSS<1:0>		<1:0>	249
DAC3CON0	EN	_	OE1	OE2	PSS	<1:0>	NSS	<1:0>	244
DAC4CON0	EN	_	OE1	OE2		<1:0>		<1:0>	244
DAC7CON0	EN	_	OE1	OE2		<1:0>		<1:0>	244
DAC8CON0 ⁽¹⁾	EN	_	OE1	OE2		<1:0>		<1:0>	244
DAC3REF						REF<4:0>	1		245
DAC4REF						REF<4:0>			245
DAC7REF						REF<4:0>			245
DAC8REF ⁽¹⁾						REF<4:0>			245
DAGOINEI									

TABLE 19-6: SUMMARY OF REGISTERS ASSOCIATED WITH COMPARATOR MODULE

 Legend:
 - = unimplemented location, read as '0'. Shaded cells are unused by the comparator module.

 Note
 1:
 PIC16LF1777/9 only.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page		
DAC2REFH		REF<9:x> (x Depends on FM bit)									
DAC5REFH		REF<9:x> (x Depends on FM bit)									
DAC6REFH ⁽¹⁾			RE	:F<9:x> (x De	pends on FM	bit)			250		
DAC1REFL			REF	= <x-1:0> (x De</x-1:0>	epends on FM	1 bit)			250		
DAC2REFL			REF	= <x-1:0> (x De</x-1:0>	epends on FM	1 bit)			250		
DAC5REFL		REF <x-1:0> (x Depends on FM bit)</x-1:0>									
DAC6REFL ⁽¹⁾	REF <x-1:0> (x Depends on FM bit)</x-1:0>								250		
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	132		
PIE2	OSFIE	C2IE	C1IE	COG1IE	BCL1IE	C4IE	C3IE	CCP2IE	134		
PIE5	CCP8IE ⁽¹⁾	CCP7IE	COG4IE ⁽¹⁾	COG3IE	C8IE ⁽¹⁾	C7IE ⁽¹⁾	C6IE	C5IE	137		
PIR2	OSFIF	C2IF	C1IF	COG1IF	BCL1IF	C4IF	C3IF	CCP2IF	140		
PIR5	CCP8IF ⁽¹⁾	CCP7IF	COG4IF ⁽¹⁾	COG3IF	C8IF ⁽¹⁾	C7IF ⁽¹⁾	C6IF	C5IF	143		
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	176		
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	181		
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	186		

TABLE 19-4: SUMMARY OF REGISTERS ASSOCIATED WITH COMPARATOR MODULE (CONT.)

Legend: — = unimplemented location, read as '0'. Shaded cells are unused by the comparator module.

Note 1: PIC16LF1777/9 only.

20.0 ZERO-CROSS DETECTION (ZCD) MODULE

The ZCD module detects when an A/C signal crosses through the ground potential. The actual zero-crossing threshold is the zero-crossing reference voltage, ZCPINV, which is typically 0.75V above ground.

The connection to the signal to be detected is through a series current limiting resistor. The module applies a current source or sink to the ZCD pin to maintain a constant voltage on the pin, thereby preventing the pin voltage from forward biasing the ESD protection diodes. When the applied voltage is greater than the reference voltage, the module sinks current. When the applied voltage is less than the reference voltage, the module sources current. The current source and sink action keeps the pin voltage. The ZCD module is shown in the simplified block diagram Figure 20-2.

The ZCD module is useful when monitoring an AC waveform for, but not limited to, the following purposes:

- A/C period measurement
- · Accurate long term time measurement
- · Dimmer phase delayed drive
- Low EMI cycle switching

FIGURE 20-2: SIMPLIFIED ZCD BLOCK DIAGRAM

20.1 External Resistor Selection

The ZCD module requires a current limiting resistor in series with the external voltage source. The impedance and rating of this resistor depends on the external source peak voltage. Select a resistor value that will drop all of the peak voltage when the current through the resistor is nominally 300 μ A. Refer to Equation 20-1 and Figure 20-1. Make sure that the ZCD I/O pin internal weak pull-up is disabled so it does not interfere with the current source and sink.

EQUATION 20-1: EXTERNAL RESISTOR

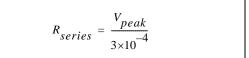
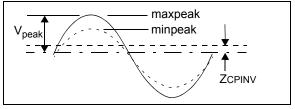
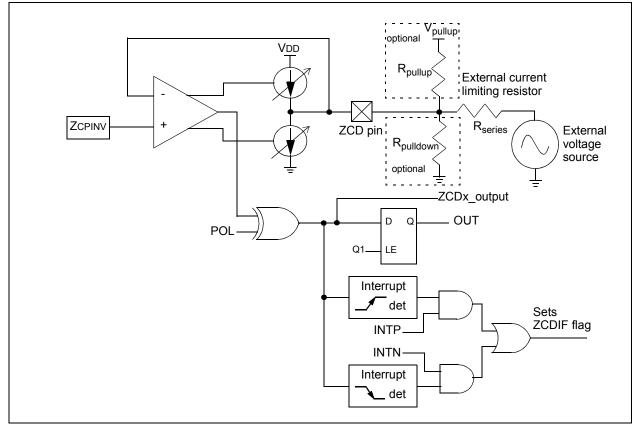


FIGURE 20-1: EXTERNAL VOLTAGE





20.2 ZCD Logic Output

The ZCD module includes a Status bit, which can be read to determine whether the current source or sink is active. The OUT bit of the ZCDCON register is set when the current sink is active, and cleared when the current source is active. The OUT bit is affected by the polarity bit.

20.3 ZCD Logic Polarity

The POL bit of the ZCDxCON register inverts the OUT bit relative to the current source and sink output. When the POL bit is set, a OUT high indicates that the current source is active, and a low output indicates that the current sink is active.

The POL bit affects the ZCD interrupts. See **Section 20.4 "ZCD Interrupts**".

20.4 ZCD Interrupts

An interrupt will be generated upon a change in the ZCD logic output when the appropriate interrupt enables are set. A rising edge detector and a falling edge detector are present in the ZCD for this purpose.

The ZCDIF bit of the PIR3 register will be set when either edge detector is triggered and its associated enable bit is set. The INTP enables rising edge interrupts and the INTN bit enables falling edge interrupts. Both are located in the ZCDxCON register.

To fully enable the interrupt, the following bits must be set:

- · ZCDIE bit of the PIE3 register
- INTP bit of the ZCDxCON register (for a rising edge detection)
- INTN bit of the ZCDxCON register (for a falling edge detection)
- · PEIE and GIE bits of the INTCON register

Changing the POL bit will cause an interrupt, regardless of the level of the EN bit.

The ZCDIF bit of the PIR3 register must be cleared in software as part of the interrupt service. If another edge is detected while this flag is being cleared, the flag will still be set at the end of the sequence.

20.5 Correcting for ZCPINV Offset

The actual voltage at which the ZCD switches is the reference voltage at the non-inverting input of the ZCD op amp. For external voltage source waveforms other than square waves this voltage offset from zero causes the zero-cross event to occur either too early or too late.

20.5.1 CORRECTION BY AC COUPLING

When the external voltage source is sinusoidal, the effects of the ZCPINV offset can be eliminated by isolating the external voltage source from the ZCD pin with a capacitor in addition to the voltage reducing resistor. The capacitor will cause a phase shift resulting in the ZCD output switch in advance of the actual zero-crossing event. The phase shift will be the same for both rising and falling zero crossings, which can be compensated for by either delaying the CPU response to the ZCD switch by a timer or other means, or selecting a capacitor value large enough that the phase shift is negligible.

To determine the series resistor and capacitor values for this configuration, start by computing the impedance, Z, to obtain a peak current of 300 μ A. Next, arbitrarily select a suitably large non-polar capacitor and compute its reactance, X_C, at the external voltage source frequency. Finally, compute the series resistor, capacitor peak voltage, and phase shift by the formulas shown in Equation 20-2.

EQUATION 20-2: R-C CALCULATIONS

$$V_{\text{peak}} = \text{external voltage source peak voltage}$$

$$f = \text{external voltage source frequency}$$

$$C = \text{series capacitor}$$

$$R = \text{series resistor}$$

$$V_{c} = \text{Peak capacitor voltage}$$

$$\phi = \text{Capacitor induced zero crossing phase}$$

$$advance in radians$$

$$T_{\phi} = \text{Time ZC event occurs before actual zero}$$

$$crossing$$

$$Z = \frac{V_{PEAK}}{3x10^{-4}}$$

$$X_{c} = \frac{1}{(2\pi fC)}$$

$$R = \sqrt{Z^{2} - X_{c}}$$

$$V_{c} = X_{c}(3x10^{-4})$$

$$\phi = Tan^{-1}(\frac{X_{c}}{R})$$

$$T_{\phi} = \frac{\phi}{(2\pi f)}$$

$$V_{rms} = 120$$

EQUATION 20-3: R-C CALCULATIONS EXAMPLE

$$V_{peak} = V_{rms} \cdot \sqrt{2} = 169.7$$

f = 60 Hz
C = 0.1 µf

$$Z = \frac{V_{peak}}{3 \times 10^{-4}} = \frac{169.7}{3 \times 10^{-4}} = 565.7 \ kOhms$$

$$X_c = \frac{1}{(2\pi fC)} = \frac{1}{(2\pi \cdot 60 \cdot 1 \cdot 10^{-7})} = 26.53 \ kOhms$$
R = 560 kOhms

$$Z_R = \sqrt{(R^2 + X_c^2)} = 560.6 \ kOhm \ (using actual resistor)$$

$$I_{peak} = \frac{V_{peak}}{Z_R} = 302.7 \cdot 10^{-6}$$

$$V_c = X_c \cdot I_{peak} = 8.0 \ V$$

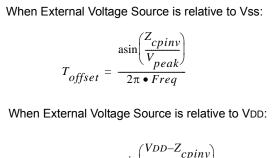
$$\phi = Tan^{-1} \left(\frac{X_c}{R}\right) = 0.047 \ radians$$

$$T_{\phi} = \frac{\phi}{(2\pi f)} = 125.6 \ \mu s$$

20.5.2 CORRECTION BY OFFSET CURRENT

When the waveform is varying relative to Vss then the zero cross is detected too early as the waveform falls and too late as the waveform rises. When the waveform is varying relative to VDD then the zero cross is detected too late as the waveform rises and too early as the waveform falls. The actual offset time can be determined for sinusoidal waveforms with the corresponding equations shown in Equation 20-4.

EQUATION 20-4: ZCD EVENT OFFSET



$$T_{offset} = \frac{\operatorname{asin}\left(\frac{v_{DD-L_{cpinv}}}{v_{peak}}\right)}{2\pi \bullet Freq}$$

This offset time can be compensated for by adding a pull-up or pull-down biasing resistor to the ZCD pin. A pull-up resistor is used when the external voltage source is varying relative to Vss. A pull-down resistor is used when the voltage is varying relative to VDD. The resistor adds a bias to the ZCD pin so that the target external voltage source must go to zero to pull the pin voltage to the ZCPINV switching voltage. The pull-up or pull-down value can be determined with the equations shown in Equation 20-5.

EQUATION 20-5: ZCD PULL-UP/DOWN

When External Signal is relative to Vss:
$R_{pullup} = \frac{R_{series}(V_{pullup} - Z_{cpinv})}{Z_{cpinv}}$
When External Signal is relative to VDD:
$R_{pulldown} = \frac{R_{series}(Z_{cpinv})}{(VDD - Z_{cpinv})}$

The pull-up and pull-down resistor values are significantly affected by small variations of ZCPINV. Measuring ZCPINV can be difficult, especially when the waveform is relative to VDD. However, by combining Equation 20-4 and Equation 20-5 the resistor value can be determined from the time difference between the ZCDOUT high and low periods. Note that the time difference, ΔT , is 4^*T_{offset} . The equation for determining the pull-up and pull-down resistor values from the high and low ZCDOUT periods is shown in Equation 20-6. The ZCDOUT signal can be directly observed on a pin by routing the ZCDOUT signal through one of the CLCs.

EQUATION 20-6:

$$R = R_{series} \left(\frac{V_{bias}}{V_{peak} \left(sin \left(\pi Freq \frac{(\Delta T)}{2} \right) \right)} - 1 \right)$$

R is pull-up or pull-down resistor

 $V_{\text{bias}} \text{ is } V_{\text{pullup}}$ when R is pull-up or VDD when R is pull-down

 ΔT is the ZCDOUT high and low period difference

20.6 Handling Vpeak variations

If the peak amplitude of the external voltage is expected to vary then the series resistor must be selected to keep the ZCD current source and sink below the design maximum range of ± 600 µA for the maximum expected voltage and high enough to be detected accurately at the minimum peak voltage. A general rule of thumb is that the maximum peak voltage can be no more than six times the minimum peak voltage. To ensure that the maximum current does not exceed ± 600 µA and the minimum is at least \pm 100 μ A, compute the series resistance as shown in Equation 20-7. The compensating pull-up for this series resistance can be determined with Equation 20-5 because the pull-up value is independent from the peak voltage.

EQUATION 20-7: SERIES R FOR V RANGE

$$R_{series} = \frac{V_{maxpeak} + V_{minpeak}}{7 \times 10^{-4}}$$

20.7 Operation During Sleep

The ZCD current sources and interrupts are unaffected by Sleep.

20.8 Effects of a Reset

The ZCD circuit can be configured to default to the active or inactive state on Power-on-Reset (POR). When the ZCD Configuration bit is cleared, the ZCD circuit will be active at POR. When the ZCD Configuration bit is set, the ZCDEN bit of the ZCDCON register must be set to enable the ZCD module.

20.9 Register Definitions: ZCD Control

Long bit name prefixes for the zero-cross detect peripheral is shown in Table 20-1. Refer to **1.1.2.2** "Long Bit Names" for more information

TABLE 20-1:

Peripheral	Bit Name Prefix
ZCD1	ZCD1

REGISTER 20-1: ZCDxCON: ZERO-CROSS DETECTION CONTROL REGISTER

			R/W-0/0	U-0	U-0	R/W-0/0	R/W-0/0
EN	_	OUT	POL	—	-	INTP	INTN
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = value depends on configuration bits

bit 7	 EN: Zero-Cross Detection Enable bit⁽¹⁾ 1 = Zero-cross detect is enabled. ZCD pin is forced to output to source and sink current. 0 = Zero-cross detect is disabled. ZCD pin operates according to PPS and TRIS controls.
bit 6	Unimplemented: Read as '0'
bit 5	OUT: Zero-Cross Detection Logic Level bit POL bit = 0: 1 = ZCD pin is sinking current 0 = ZCD pin is sourcing current POL bit = 1: 1 = ZCD pin is sourcing current 0 = ZCD pin is sourcing current 0 = ZCD pin is sourcing current 0 = ZCD pin is sourcing current
bit 4	 POL: Zero-Cross Detection Logic Output Polarity bit 1 = ZCD logic output is inverted 0 = ZCD logic output is not inverted
bit 3-2	Unimplemented: Read as '0'
bit 1	 INTP: Zero-Cross Positive Edge Interrupt Enable bit 1 = ZCDIF bit is set on low-to-high OUT transition 0 = ZCDIF bit is unaffected by low-to-high OUT transition
bit 0	 INTN: Zero-Cross Negative Edge Interrupt Enable bit 1 = ZCDIF bit is set on high-to-low OUT transition 0 = ZCDIF bit is unaffected by high-to-low OUT transition
Note 1:	The EN bit has no effect when the $\overline{\text{ZCD}}$ Configuration bit is cleared.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on page
PIE3	_	_	COG2IE	ZCDIE	CLC4IE	CLC3IE	CLC2IE	CLC1IE	135
PIR3	_	_	COG2IF	ZCDIF	CLC4IF	CLC3IF	CLC2IF	CLC1IF	141
ZCD1CON	EN	_	OUT	POL		_	INTP	INTN	270

TABLE 20-2: SUMMARY OF REGISTERS ASSOCIATED WITH THE ZCD MODULE

Legend: — = unimplemented, read as '0'. Shaded cells are unused by the ZCD module.

TABLE 20-3: SUMMARY OF CONFIGURATION WORD WITH THE ZCD MODULE

Name	Bits	Bit -/7	Bit -/6	Bit 13/5	Bit 12/4	Bit 11/3	Bit 10/2	Bit 9/1	Bit 8/0	Register on Page
CONFIG2	13:8	_	_	LVP	DEBUG	LPBOR	BORV	STVREN	PLLEN	97
	7:0	ZCD	_	_	_	_	PPS1WAY	WRT	<1:0>	

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by the ZCD module.

21.0 TIMER0 MODULE

The Timer0 module is an 8-bit timer/counter with the following features:

- 8-bit timer/counter register (TMR0)
- 8-bit prescaler (independent of Watchdog Timer)
- Programmable internal or external clock source
- Programmable external clock edge selection
- · Interrupt-on-overflow
- TMR0 can be used to gate Timer1

Figure 21-1 is a block diagram of the Timer0 module.

21.1 Timer0 Operation

The Timer0 module can be used as either an 8-bit timer or an 8-bit counter.

21.1.1 8-BIT TIMER MODE

The Timer0 module will increment every instruction cycle, if used without a prescaler. 8-bit Timer mode is selected by clearing the TMR0CS bit of the OPTION_REG register.

When TMR0 is written, the increment is inhibited for two instruction cycles immediately following the write.

Note: The value written to the TMR0 register can be adjusted, in order to account for the two instruction cycle delay when TMR0 is written.

FIGURE 21-1: BLOCK DIAGRAM OF THE TIMER0

Fosc/4 Data Bus Set TMR0IF **T0CKIPPS** 0 ł 8 TOCKI 1 Sync PPS 1 TMR0 2 TCY Timer0 overflow 0 TMR0SE TMR0CS 8-bit Prescaler PSA PS<2:0>

21.1.2 8-BIT COUNTER MODE

In 8-Bit Counter mode, the Timer0 module will increment on every rising or falling edge of the T0CKI pin.

8-Bit Counter mode using the T0CKI pin is selected by setting the TMR0CS bit in the OPTION_REG register to '1'.

The rising or falling transition of the incrementing edge for either input source is determined by the TMR0SE bit in the OPTION_REG register.

21.1.3 SOFTWARE PROGRAMMABLE PRESCALER

A software programmable prescaler is available for exclusive use with Timer0. The prescaler is enabled by clearing the PSA bit of the OPTION_REG register.

Note:	The Watchdog Timer (WDT) uses its own
	independent prescaler.

There are eight prescaler options for the Timer0 module ranging from 1:2 to 1:256. The prescale values are selectable via the PS<2:0> bits of the OPTION_REG register. In order to have a 1:1 prescaler value for the Timer0 module, the prescaler must be disabled by setting the PSA bit of the OPTION_REG register.

The prescaler is not readable or writable. All instructions writing to the TMR0 register will clear the prescaler.

21.1.4 TIMER0 INTERRUPT

Timer0 will generate an interrupt when the TMR0 register overflows from FFh to 00h. The TMR0IF interrupt flag bit of the INTCON register is set every time the TMR0 register overflows, regardless of whether or not the Timer0 interrupt is enabled. The TMR0IF bit can only be cleared in software. The Timer0 interrupt enable is the TMR0IE bit of the INTCON register.

Note:	The Timer0 interrupt cannot wake the
	processor from Sleep since the timer is
	frozen during Sleep.

21.1.5 8-BIT COUNTER MODE SYNCHRONIZATION

When in 8-Bit Counter mode, the incrementing edge on the T0CKI pin must be synchronized to the instruction clock. Synchronization can be accomplished by sampling the prescaler output on the Q2 and Q4 cycles of the instruction clock. The high and low periods of the external clocking source must meet the timing requirements as shown in Table 36-12: Timer0 and Timer1 External Clock Requirements.

21.1.6 OPERATION DURING SLEEP

Timer0 cannot operate while the processor is in Sleep mode. The contents of the TMR0 register will remain unchanged while the processor is in Sleep mode.

21.2 Register Definitions: Option Register

REGISTER 21-1: OPTION_REG: OPTION REGISTER

R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1
WPUEN	INTEDG	TMR0CS	TMR0SE	PSA		PS<2:0>	
oit 7							bit (
Legend:							
R = Readable	bit	W = Writable	bit	U = Unimpler	mented bit, read	d as '0'	
u = Bit is unch	anged	x = Bit is unkr	nown	-n/n = Value :	at POR and BC	R/Value at all o	other Resets
1' = Bit is set	0	'0' = Bit is clea	ared				
bit 7		ak Pull-Up Ena					
	•	oull-ups are dis -ups are enabl	• •		,		
bit 6	•	rrupt Edge Sel	•	al WFUX later	values		
		on rising edge					
	•	on falling edge	•				
oit 5	•	ner0 Clock Sou	•				
	1 = Transition	on T0CKI pin					
	0 = Internal in	struction cycle	clock (Fosc/4	4)			
bit 4	TMR0SE: Tim	er0 Source Ec	lge Select bit				
		on high-to-lov					
	0 = Increment	on low-to-high	n transition on	T0CKI pin			
bit 3		er Assignment					
		is not assigned to					
bit 2-0	PS<2:0>: Pre	scaler Rate Se	elect bits				
	Bit V	/alue Timer0	Rate				
		00 1:2					
		01 1:4 10 1:8					
		10 1:10 11 1:10					
		00 1:3					
		01 1:6					
	1	10 1:1	28				
	1	11 1:2	56				

TABLE 21-1: SUMMARY OF REGISTERS ASSOCIATED WITH TIMER0

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	132
OPTION_REG	WPUEN	INTEDG	TMR0CS	TMR0SE	PSA		PS<2:0>		274
TMR0	Timer0 Mo	Timer0 Module Register					272*		
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	176

Legend: — = Unimplemented location, read as '0'. Shaded cells are not used by the Timer0 module.

* Page provides register information.

22.0 TIMER1/3/5 MODULE WITH GATE CONTROL

The Timer1 module is a 16-bit timer/counter with the following features:

- 16-bit timer/counter register pair (TMR1H:TMR1L)
- Programmable internal or external clock source
- · 2-bit prescaler
- · Dedicated 32 kHz oscillator circuit
- · Optionally synchronized comparator out
- · Multiple Timer1 gate (count enable) sources
- Interrupt-on-overflow
- Wake-up on overflow (external clock, Asynchronous mode only)
- Time base for the Capture/Compare function
- Auto-conversion Trigger (with CCP)

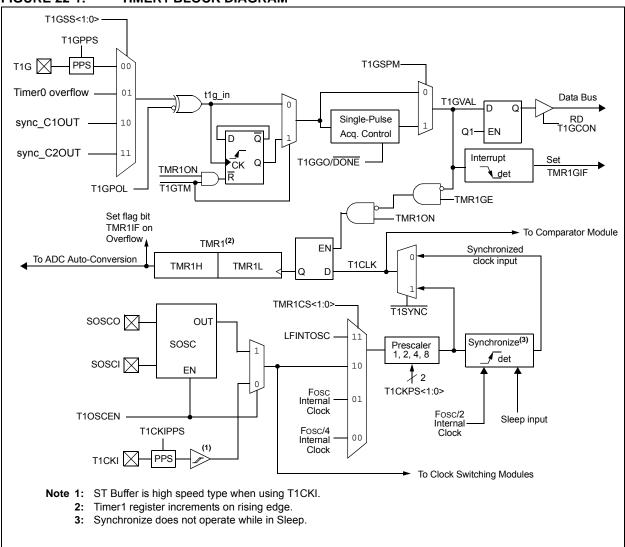
- Selectable Gate Source Polarity
- Gate Toggle mode
- · Gate Single-pulse mode
- Gate Value Status
- Gate Event Interrupt

Figure 22-1 is a block diagram of the Timer1 module.

This device has three instances of Timer1 type modules. They include:

- Timer1
- Timer3
- Timer5

Note: All references to Timer1 and Timer1 Gate apply to Timer3 and Timer5.





22.1 Timer1 Operation

The Timer1 module is a 16-bit incrementing counter which is accessed through the TMR1H:TMR1L register pair. Writes to TMR1H or TMR1L directly update the counter.

When used with an internal clock source, the module is a timer and increments on every instruction cycle. When used with an external clock source, the module can be used as either a timer or counter and increments on every selected edge of the external source.

Timer1 is enabled by configuring the ON and GE bits in the T1CON and T1GCON registers, respectively. Table 22-1 displays the Timer1 enable selections.

TABLE 22-1:	TIMER1 ENABLE
	SELECTIONS

TMR10N	TMR1GE	Timer1 Operation
0	0	Off
0	1	Off
1	0	Always On
1	1	Count Enabled

22.2 Clock Source Selection

The CS<1:0> and OSCEN bits of the T1CON register are used to select the clock source for Timer1. Table 22-2 displays the clock source selections.

22.2.1 INTERNAL CLOCK SOURCE

When the internal clock source is selected, the TMR1H:TMR1L register pair will increment on multiples of Fosc as determined by the Timer1 prescaler.

When the Fosc internal clock source is selected, the Timer1 register value will increment by four counts every instruction clock cycle. Due to this condition, a 2 LSB error in resolution will occur when reading the Timer1 value. To utilize the full resolution of Timer1, an asynchronous input signal must be used to gate the Timer1 clock input.

The following asynchronous sources may be used:

- Asynchronous event on the T1G pin to Timer1
 gate
- C1 or C2 comparator input to Timer1 gate

22.2.2 EXTERNAL CLOCK SOURCE

When the external clock source is selected, the Timer1 module may work as a timer or a counter.

When enabled to count, Timer1 is incremented on the rising edge of the external clock input T1CKI, which can be synchronized to the microcontroller system clock or can run asynchronously.

When used as a timer with a clock oscillator, an external 32.768 kHz crystal can be used in conjunction with the dedicated internal oscillator circuit.

- Note: In Counter mode, a falling edge must be registered by the counter prior to the first incrementing rising edge after any one or more of the following conditions:
 - · Timer1 enabled after POR
 - Write to TMR1H or TMR1L
 - Timer1 is disabled
 - Timer1 is disabled (TMR1ON = 0) when T1CKI is high then Timer1 is enabled (TMR1ON=1) when T1CKI is low.

TABLE 22-2: CLOCK SOURCE SELECTIONS

TMR1CS<1:0>	T10SCEN	Clock Source
11	х	LFINTOSC
10	0	External Clocking on T1CKI Pin
01	x	System Clock (Fosc)
00	x	Instruction Clock (Fosc/4)

22.3 Timer1 Prescaler

Timer1 has four prescaler options allowing 1, 2, 4 or 8 divisions of the clock input. The CKPS bits of the T1CON register control the prescale counter. The prescale counter is not directly readable or writable; however, the prescaler counter is cleared upon a write to TMR1H or TMR1L.

22.4 Timer1 (Secondary) Oscillator

A dedicated low-power 32.768 kHz oscillator circuit is built-in between pins SOSCI (input) and SOSCO (amplifier output). This internal circuit is to be used in conjunction with an external 32.768 kHz crystal.

The oscillator circuit is enabled by setting the OSCEN bit of the T1CON register. The oscillator will continue to run during Sleep.

Note: The oscillator requires a start-up and stabilization time before use. Thus, OSCEN should be set and a suitable delay observed prior to using Timer1. A suitable delay similar to the OST delay can be implemented in software by clearing the TMR1IF bit then presetting the TMR1H:TMR1L register pair to FC00h. The TMR1IF flag will be set when 1024 clock cycles have elapsed, thereby indicating that the oscillator is running and reasonably stable.

22.5 Timer1 Operation in Asynchronous Counter Mode

If the control bit SYNC of the T1CON register is set, the external clock input is not synchronized. The timer increments asynchronously to the internal phase clocks. If the external clock source is selected then the timer will continue to run during Sleep and can generate an interrupt-on-overflow, which will wake-up the processor. However, special precautions in software are needed to read/write the timer (see Section 22.5.1 "Reading and Writing Timer1 in Asynchronous Counter Mode").

Note:	When switching from synchronous to
	asynchronous operation, it is possible to
	skip an increment. When switching from
	asynchronous to synchronous operation,
	it is possible to produce an additional
	increment.

22.5.1 READING AND WRITING TIMER1 IN ASYNCHRONOUS COUNTER MODE

Reading TMR1H or TMR1L while the timer is running from an external asynchronous clock will ensure a valid read (taken care of in hardware). However, the user should keep in mind that reading the 16-bit timer in two 8-bit values itself, poses certain problems, since the timer may overflow between the reads.

For writes, it is recommended that the user simply stop the timer and write the desired values. A write contention may occur by writing to the timer registers, while the register is incrementing. This may produce an unpredictable value in the TMR1H:TMR1L register pair.

22.6 Timer1 Gate

Timer1 can be configured to count freely or the count can be enabled and disabled using Timer1 gate circuitry. This is also referred to as Timer1 Gate Enable.

Timer1 gate can also be driven by multiple selectable sources.

22.6.1 TIMER1 GATE ENABLE

The Timer1 Gate Enable mode is enabled by setting the GE bit of the T1GCON register. The polarity of the Timer1 Gate Enable mode is configured using the GPOL bit of the T1GCON register.

When Timer1 Gate Enable mode is enabled, Timer1 will increment on the rising edge of the Timer1 clock source. When Timer1 Gate Enable mode is disabled, no incrementing will occur and Timer1 will hold the current count. See Figure 22-3 for timing details.

TABLE 22-3: TIMER1 GATE ENABLE SELECTIONS

T1CLK	T1GPOL	T1G	Timer1 Operation
\uparrow	1	1	Counts
\uparrow	1	0	Holds Count
\uparrow	0	1	Holds Count
\uparrow	0	0	Counts

22.6.2 TIMER1 GATE SOURCE SELECTION

Timer1 gate source selections are shown in Table 22-4. Source selection is controlled by the T1GSS bits of the T1GCON register. The polarity for each available source is also selectable. Polarity selection is controlled by the T1GPOL bit of the T1GCON register.

TABLE 22-4 :	TIMER1	GATE SOURCE	S
---------------------	--------	-------------	---

T1GSS	Timer1 Gate Source
11	Comparator 2 Output sync_C2OUT (optionally Timer1 synchronized output)
10	Comparator 1 Output sync_C1OUT (optionally Timer1 synchronized output)
01	Overflow of Timer0 (TMR0 increments from FFh to 00h)
00	Timer1 Gate Pin

22.6.2.1 T1G Pin Gate Operation

The T1G pin is one source for Timer1 gate control. It can be used to supply an external source to the Timer1 gate circuitry.

22.6.2.2 Timer0 Overflow Gate Operation

When Timer0 increments from FFh to 00h, a low-to-high pulse will automatically be generated and internally supplied to the Timer1 gate circuitry.

22.6.2.3 Comparator C1 Gate Operation

The output resulting from a Comparator 1 operation can be selected as a source for Timer1 gate control. The Comparator 1 output (sync_C1OUT) can be synchronized to the Timer1 clock or left asynchronous. For more information see Section 19.4.1 "Comparator Output Synchronization".

22.6.2.4 Comparator C2 Gate Operation

The output resulting from a Comparator 2 operation can be selected as a source for Timer1 gate control. The Comparator 2 output (sync_C2OUT) can be synchronized to the Timer1 clock or left asynchronous. For more information see Section 19.4.1 "Comparator Output Synchronization".

22.6.3 TIMER1 GATE TOGGLE MODE

When Timer1 Gate Toggle mode is enabled, it is possible to measure the full-cycle length of a Timer1 gate signal, as opposed to the duration of a single level pulse.

The Timer1 gate source is routed through a flip-flop that changes state on every incrementing edge of the signal. See Figure 22-4 for timing details.

Timer1 Gate Toggle mode is enabled by setting the T1GTM bit of the T1GCON register. When the T1GTM bit is cleared, the flip-flop is cleared and held clear. This is necessary in order to control which edge is measured.

Note:	Enabling Toggle mode at the same time
	as changing the gate polarity may result in
	indeterminate operation.

22.6.4 TIMER1 GATE SINGLE-PULSE MODE

When Timer1 Gate Single-Pulse mode is enabled, it is possible to capture a single-pulse gate event. Timer1 Gate Single-Pulse mode is first enabled by setting the T1GSPM bit in the T1GCON register. Next, the T1GGO/DONE bit in the T1GCON register must be set. The Timer1 will be fully enabled on the next incrementing edge. On the next trailing edge of the pulse, the T1GGO/DONE bit will automatically be cleared. No other gate events will be allowed to increment Timer1 until the T1GGO/DONE bit is once again set in software. See Figure 22-5 for timing details.

If the Single-Pulse Gate mode is disabled by clearing the T1GSPM bit in the T1GCON register, the T1GGO/DONE bit should also be cleared.

Enabling the Toggle mode and the Single-Pulse mode simultaneously will permit both sections to work together. This allows the cycle times on the Timer1 gate source to be measured. See Figure 22-6 for timing details.

22.6.5 TIMER1 GATE VALUE STATUS

When Timer1 Gate Value Status is utilized, it is possible to read the most current level of the gate control value. The value is stored in the T1GVAL bit in the T1GCON register. The T1GVAL bit is valid even when the Timer1 gate is not enabled (TMR1GE bit is cleared).

22.6.6 TIMER1 GATE EVENT INTERRUPT

When Timer1 Gate Event Interrupt is enabled, it is possible to generate an interrupt upon the completion of a gate event. When the falling edge of T1GVAL occurs, the TMR1GIF flag bit in the PIR1 register will be set. If the TMR1GIE bit in the PIE1 register is set, then an interrupt will be recognized.

The TMR1GIF flag bit operates even when the Timer1 gate is not enabled (TMR1GE bit is cleared).

22.7 **Timer1 Interrupt**

The Timer1 register pair (TMR1H:TMR1L) increments to FFFFh and rolls over to 0000h. When Timer1 rolls over, the Timer1 interrupt flag bit of the PIR1 register is set. To enable the interrupt-on-rollover, you must set these bits:

- · ON bit of the T1CON register
- TMR1IE bit of the PIE1 register
- PEIE bit of the INTCON register
- · GIE bit of the INTCON register

The interrupt is cleared by clearing the TMR1IF bit in the Interrupt Service Routine.

Note: The TMR1H:TMR1L register pair and the TMR1IF bit should be cleared before enabling interrupts.

22.8 **Timer1 Operation During Sleep**

Timer1 can only operate during Sleep when setup in Asynchronous Counter mode. In this mode, an external crystal or clock source can be used to increment the counter. To set up the timer to wake the device:

- ON bit of the T1CON register must be set
- TMR1IE bit of the PIE1 register must be set
- · PEIE bit of the INTCON register must be set
- SYNC bit of the T1CON register must be set
- · CS bits of the T1CON register must be configured
- OSCEN bit of the T1CON register must be configured

The device will wake-up on an overflow and execute the next instructions. If the GIE bit of the INTCON register is set, the device will call the Interrupt Service Routine.

Secondary oscillator will continue to operate in Sleep regardless of the SYNC bit setting.

22.9 **CCP Capture/Compare Time Base**

The CCP modules use the TMR1H:TMR1L register pair as the time base when operating in Capture or Compare mode.

In Capture mode, the value in the TMR1H:TMR1L register pair is copied into the CCPR1H:CCPR1L register pair on a configured event.

In Compare mode, an event is triggered when the value in the CCPR1H:CCPR1L register pair matches the value in the TMR1H:TMR1L register pair. This event can be an Auto-conversion Trigger.

For more information, see Section 24.0 "Capture/Compare/PWM Modules".

22.10 CCP Auto-Conversion Trigger

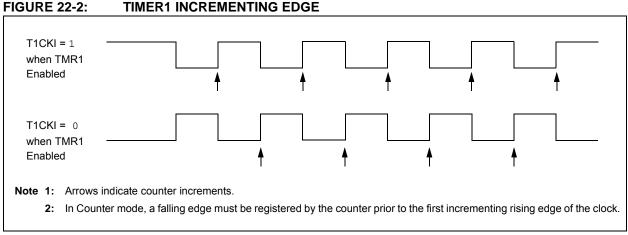
When any of the CCP's are configured to trigger an auto-conversion, the trigger will clear the TMR1H:TMR1L register pair. This auto-conversion does not cause a Timer1 interrupt. The CCP module may still be configured to generate a CCP interrupt.

In this mode of operation, the CCPR1H:CCPR1L register pair becomes the period register for Timer1.

Timer1 should be synchronized and Fosc/4 should be selected as the clock source in order to utilize the Auto-conversion Trigger. Asynchronous operation of Timer1 can cause an Auto-conversion Trigger to be missed.

In the event that a write to TMR1H or TMR1L coincides with an Auto-conversion Trigger from the CCP, the write will take precedence.

For more information, see Section 24.2.1 "Auto-Conversion Trigger".



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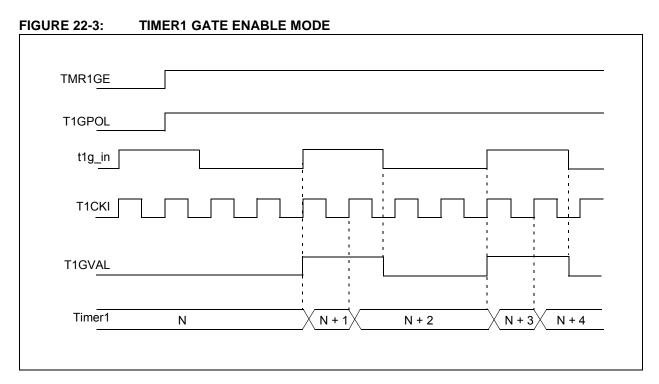


FIGURE 22-4: TIMER1 GATE TOGGLE MODE

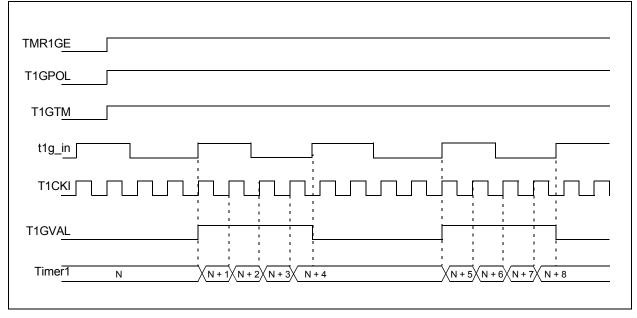


FIGURE 22-5:	TIMER1 GATE SINGLE-PULSE MODE
TMR1GE	
T1GPOL	
T1GSPM	
T1GG <u>O/</u> DONE	Cleared by hardware on falling edge of T1GVAL Counting enabled on
t1g_in	rising edge of T1G
т1СКІ	
T1GVAL	
Timer1	N N + 1 N + 2
TMR1GIF	Cleared by software Cleared by hardware on falling edge of T1GVAL

FIGURE 22-6:	TIMER1 GATE SINGLE-PULSE AND TOGGLE COMBINED MODE	
TMR1GE		
T1GPOL		
T1GSPM		
T1GTM		
T1GG <u>O/</u> DONE	Cleared by hardware o Cleared by hardware o falling edge of T1GVAL Counting enabled on rising edge of T10	n
t1g_in	rising edge of T1G	
Т1СКІ		
T1GVAL		
Timer1	N N + 1 N + 2 N + 3 N + 4	
TMR1GIF	Cleared by software falling edge of T1GVAL	

22.11 Register Definitions: Timer1 Control

Long bit name prefixes for the Timer1 peripherals are shown in Table 22-5. Refer to **Section 1.1.2.2 "Long Bit Names**" for more information

TABLE 22-5:

Peripheral	Bit Name Prefix
Timer1	T1
Timer3	Т3
Timer5	T5

REGISTER 22-1: T1CON: TIMER1 CONTROL REGISTER

R/W-0/u	R/W-0/u	R/W-0/u	R/W-0/u	R/W-0/u	R/W-0/u	U-0	R/W-0/u
CS<1:0>		CKPS<1:0>		OSCEN ⁽¹⁾	SYNC	—	ON
bit 7							bit 0

R = Readable	bit	W = Writable bit	U = Unimplemented bit, read as '0'					
u = Bit is unchanged x = Bit is unknown			-n/n = Value at POR and BOR/Value at all other Resets					
'1' = Bit is set		'0' = Bit is cleared						
bit 7-6	CS<1:0>	: Timer1 Clock Source Select	bits					
	11 = LFINTOSC							
	10 = Timer1 clock source is pin or oscillator: ⁽¹⁾							
	<u>If T10SCEN = 0</u> :							
	External clock from T1CKI pin (on the rising edge)							
	If T1OSCEN = 1: Crystal oscillator on SOSCI/SOSCO pins							
	01 = Timer1 clock source is system clock (Fosc)							
		er1 clock source is instruction						
bit 5-4	CKPS<1:0>: Timer1 Input Clock Prescale Select bits							
	11 = 1:8 Prescale value							
	10 = 1:4 Prescale value							
	01 = 1:2 Prescale value 00 = 1:1 Prescale value							
h # 0			⊨ :a(1)					
bit 3	OSCEN: LP Oscillator Enable Control bit ⁽¹⁾							
	 1 = Dedicated secondary oscillator circuit enabled 0 = Dedicated secondary oscillator circuit disabled 							
bit 2		mer1 Synchronization Contro						
	1 = Do not synchronize asynchronous clock input							
			input with system clock (Fosc)					
bit 1	Unimplemented: Read as '0'							
bit 0	ON: Timer1 On bit							
	1 = Enables Timer1							
	0 = Stops	s Timer1 and clears Timer1 g	ate flip-flop					

R/W-0/u	R/W-0/u	R/W-0/u	R/W-0/u	R/W/HC-0/u	R-x/x	R/W-0/u	R/W-0/u			
GE	GE GPOL GTM			GGO/ DONE	GVAL	GSS	<1:0>			
bit 7		·	•			·	bit 0			
Legend:										
R = Readable	bit	W = Writable	hit	II = Unimplem	ented hit rea	d as '0'				
u = Bit is unch		x = Bit is unkr		U = Unimplemented bit, read as '0' -n/n = Value at POR and BOR/Value at all other Resets						
'1' = Bit is set	0	'0' = Bit is clea		HC = Bit is cle						
bit 7	GE: Timer1 Gate Enable bit <u>If TMR1ON = 0</u> : This bit is ignored <u>If TMR1ON = 1</u> : 1 = Timer1 counting is controlled by the Timer1 gate function									
bit 6	 0 = Timer1 counts regardless of Timer1 gate function GPOL: Timer1 Gate Polarity bit 1 = Timer1 gate is active-high (Timer1 counts when gate is high) 0 = Timer1 gate is active-low (Timer1 counts when gate is low) 									
bit 5	GTM: Timer1 Gate Toggle Mode bit 1 = Timer1 Gate Toggle mode is enabled 0 = Timer1 Gate Toggle mode is disabled and toggle flip-flop is cleared Timer1 gate flip-flop toggles on every rising edge.									
bit 4	GSPM: Timer1 Gate Single-Pulse Mode bit 1 = Timer1 Gate Single-Pulse mode is enabled and is controlling Timer1 gate									
	0 = Timer1 Gate Single-Pulse mode is disabled									
bit 3	GGO/DONE: Timer1 Gate Single-Pulse Acquisition Status bit									
	 1 = Timer1 gate single-pulse acquisition is ready, waiting for an edge 0 = Timer1 gate single-pulse acquisition has completed or has not been started 									
bit 2	GVAL: Timer1 Gate Value Status bit									
		current state of y Timer1 Gate I		ate that could be GE)	e provided to 7	rmr1H:TMR1L				
bit 1-0	GSS<1:0>: Timer1 Gate Source Select bits									
	 11 = Comparator 2 optionally synchronized output (sync_C2OUT) 10 = Comparator 1 optionally synchronized output (sync_C1OUT) 01 = Timer0 overflow output 00 = Timer1 gate pin 									

REGISTER 22-2: T1GCON: TIMER1 GATE CONTROL REGISTER

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ANSELA	—	_	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0	177
CCPxCON	EN	_	OUT	FMT		MODE	<3:0>		319
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	132
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	CCP1IE	TMR2IE	TMR1IE	133
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	CCP1IF	TMR2IF	TMR1IF	139
TMRxH	Holding Register for the Most Significant Byte of the 16-bit TMR1/3/5 Register								275*
TMRxL	Holding Register for the Least Significant Byte of the 16-bit TMR1/3/5 Register							275*	
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	176
TxCON	CS<	1:0>	CKPS	<1:0>	OSCEN	SYNC	_	ON	283
TxGCON	GE	GPOL	GTM	GSPM	GGO/ DONE	GVAL	GSS	284	

TABLE 22-6: SUMMARY OF REGISTERS ASSOCIATED WITH TIMER1

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by the Timer1 module. * Page provides register information.

23.0 TIMER2/4/6/8 MODULE

The Timer2/4/6/8 modules are 8-bit timers that can operate as free-running period counters or in conjunction with external signals that control start, run, freeze, and reset operation in One-Shot and Monostable modes of operation. Sophisticated waveform control such as pulse density modulation are possible by combining the operation of these timers with other internal peripherals such as the comparators and CCP modules. Features of the timer include:

- · 8-bit timer register
- 8-bit period register
- · Selectable external hardware timer Resets
- Programmable prescaler (1:1 to 1:128)
- Programmable postscaler (1:1 to 1:16)
- Selectable synchronous/asynchronous operation
- · Alternate clock sources
- · Interrupt-on-period

- Three modes of operation:
 - Free Running Period
 - One-shot
 - Monostable

See Figure 23-1 for a block diagram of Timer2. See Figure 23-2 for the clock source block diagram.

Note: Three identical Timer2 modules are implemented on this device. The timers are named Timer2, Timer4, Timer6, and Timer8. All references to Timer2 apply as well to Timer4, Timer6 and Timer8. All references to T2PR apply as well to T4PR, T6PR and T8PR.

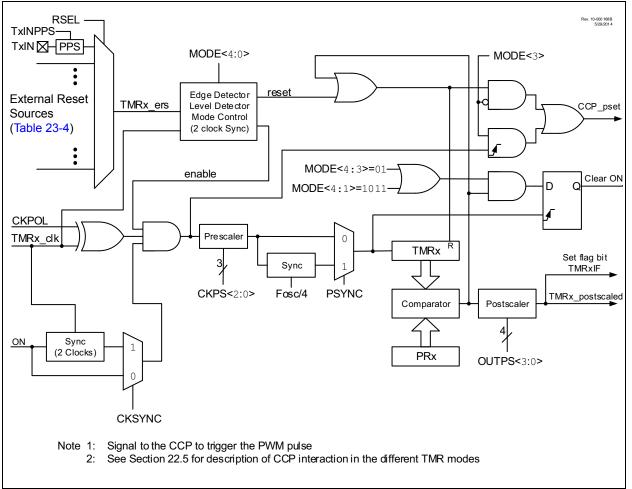
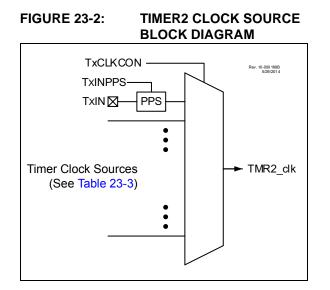


FIGURE 23-1: TIMER2 BLOCK DIAGRAM



23.1 Timer2 Operation

Timer2 operates in three major modes:

- · Free Running Period
- One-shot
- Monostable

Within each mode there are several options for starting, stopping, and reset. Table 23-1 lists the options.

In all modes, the TMR2 count register is incremented on the rising edge of the clock signal from the programmable prescaler. When TMR2 equals T2PR, a high level is output to the postscaler counter. TMR2 is cleared on the next clock input.

An external signal from hardware can also be configured to gate the timer operation or force a TMR2 count Reset. In Gate modes the counter stops when the gate is disabled and resumes when the gate is enabled. In Reset modes the TMR2 count is reset on either the level or edge from the external source.

The TMR2 and T2PR registers are both directly readable and writable. The TMR2 register is cleared and the T2PR register initializes to FFh on any device Reset. Both the prescaler and postscaler counters are cleared on the following events:

- · a write to the TMR2 register
- a write to the T2CON register
- · any device Reset
- External Reset Source event that resets the timer.

Note: TMR2 is not cleared when T2CON is written.

23.1.1 FREE RUNNING PERIOD MODE

The value of TMR2 is compared to that of the Period register, T2PR, on each clock cycle. When the two values match, the comparator resets the value of TMR2 to 00h on the next cycle and increments the output postscaler counter. When the postscaler count equals the value in the OUTPS<4:0> bits of the TMRxCON1 register then a one clock period wide pulse occurs on the TMR2_postscaled output, and the postscaler count is cleared.

23.1.2 ONE-SHOT MODE

The One-Shot mode is identical to the Free Running Period mode except that the ON bit is cleared and the timer is stopped when TMR2 matches T2PR and will not restart until the T2ON bit is cycled off and on. Postscaler OUTPS<4:0> values other than 0 are meaningless in this mode because the timer is stopped at the first period event and the postscaler is reset when the timer is restarted.

23.1.3 MONOSTABLE MODE

Monostable modes are similar to One-Shot modes except that the ON bit is not cleared and the timer can be restarted by an external Reset event.

23.2 PRx Period Register

The PRx period register is double buffered, software reads and writes the PRx register. However, the timer uses a buffered PRx register for operation. Software does not have direct access to the buffered PRx register. The contents of the PRx register is transferred to the buffer by any of the following events:

- · A write to the TMRx register
- A write to the TMRxCON register
- When TMRx = PRx buffer and the prescaler rolls over
- · An external Reset event

23.3 Timer2 Output

The Timer2 module's primary output is TMR2_postscaled, which pulses for a single TMR2_clk period when the postscaler counter matches the value in the OUTPS bits of the TMR2xCON register. The T2PR postscaler is incremented each time the TMR2 value matches the T2PR value. This signal can be selected as an input to several other input modules:

- The ADC module, as an Auto-conversion Trigger
- · COG, as an auto-shutdown source

In addition, the Timer2 is also used by the CCP module for pulse generation in PWM mode. Both the actual TMR2 value as well as other internal signals are sent to the CCP module to properly clock both the period and pulse width of the PWM signal. See **Section 24.6** "**CCP/PWM Clock Selection**" for more details on setting up Timer2 for use with the CCP, as well as the timing diagrams in **Section 23.6** "**Operation Examples**" for examples of how the varying Timer2 modes affect CCP PWM output.

23.4 External Reset Sources

In addition to the clock source, the Timer2 also takes in an external Reset source. This external Reset source is selected for Timer2, Timer4, Timer6 and Timer8 with the T2RST, T4RST, T6RST and T8RST registers, respectively. This source can control starting and stopping of the timer, as well as resetting the timer, depending on which mode the timer is in. The mode of the timer is controlled by the MODE<4:0> bits of the TMRxHLT register. Edge-Triggered modes require six Timer clock periods between external triggers. Level-Triggered modes require the triggering level to be at least three Timer clock periods long. External triggers are ignored while in Debug Freeze mode.

Mode -	MODE	E<4:0>	Output	Onenstien	Timer Control			
	<4:3>	<2:0>	Operation	Operation	Start	Reset	Stop	
		000		Software gate (Figure 23-4)	ON = 1	—	ON = 0	
Free Running Period		001	Period Pulse	Hardware gate, active-high (Figure 23-5)	ON = 1 and TMRx_ers = 1	_	ON = 0 or TMRx_ers = 0	
		010	i uise	Hardware gate, active-low	ON = 1 and TMRx_ers = 0	—	ON = 0 or TMRx_ers = 1	
	0.0	011		Rising or falling edge Reset		TMRx_ers		
	00	100	Period	Rising edge Reset (Figure 23-6)		TMRx_ers ↑	ON = 0	
		101	Pulse	Falling edge Reset	-	TMRx_ers ↓		
		110	with Hardware	Low level Reset	ON = 1	TMRx_ers = 0	ON = 0 or TMRx_ers =	
		111	Reset	High level Reset (Figure 23-7)	-	TMRx_ers = 1	ON = 0 or TMRx_ers = 2	
		000	One-shot	Software start (Figure 23-8)	ON = 1	_		
		001 Ed	Edge	Rising edge start (Figure 23-9)	ON = 1 and TMRx_ers ↑	_		
	01	010	triggered start	Falling edge start	ON = 1 and TMRx_ers ↓	_		
		011	(Note 1)	Any edge start	ON = 1 and TMRx_ers	_	ON = 0 or	
One-shot		100	Edge triggered start and hardware Reset (Note 1)	Rising edge start and Rising edge Reset (Figure 23-10)	ON = 1 and TMRx_ers ↑	TMRx_ers ↑	Next clock after TMRx = PRx (Note 2)	
		101		Falling edge start and Falling edge Reset	ON = 1 and TMRx_ers ↓	TMRx_ers ↓		
		110		Rising edge start and Low level Reset (Figure 23-11)	ON = 1 and TMRx_ers ↑	TMRx_ers = 0		
		111		Falling edge start and High level Reset	ON = 1 and TMRx_ers ↓	TMRx_ers = 1		
		000		Rese	erved			
		001	Edge	Rising edge start (Figure 23-12)	ON = 1 and TMRx_ers ↑	—	ON = 0 or	
Mono-stable		010	triggered start	Falling edge start	ON = 1 and TMRx_ers ↓	—	Next clock after	
		011	(Note 1)	Any edge start	ON = 1 and TMRx_ers	—	TMRx = PRx (Note 3)	
Reserved	10	100		Rese	erved			
Reserved		101		Rese	erved			
One-shot		110	Level triggered	High level start and Low level Reset (Figure 23-13)	ON = 1 and TMRx_ers = 1	TMRx_ers = 0	ON = 0 or	
		111	start and hardware Reset	Low level start & High level Reset	ON = 1 and TMRx_ers = 0	TMRx_ers = 1	Held in Rese (Note 2)	
Reserved	11	xxx		Rese	erved			

TABLE 23-1: TIMER2 OPERATING MODES

Note 1: If ON = 0 then an edge is required to restart the timer after ON = 1.

2: When TMRx = PRx then the next clock clears ON and stops TMRx at 00h.

3: When TMRx = PRx then the next clock stops TMRx at 00h but does not clear ON.

23.5 Timer2 Interrupt

Timer2 can also generate a device interrupt. The interrupt is generated when the postscaler counter matches one of 16 postscale options (from 1:1 through 1:16), which are selected with the postscaler control bits, OUTPS<3:0> of the T2CON register. The interrupt is enabled by setting the TMR2IE interrupt enable bit of the PIE1 register. Interrupt timing is illustrated in Figure 23-3.

FIGURE 23-3: TIMER2 PRESCALER, POSTSCALER, AND INTERRUPT TIMING DIAGRAM

	Rev. 10.000016A 4/72016
CKPS	0b010
PRx	1
OUTPS	0b0001
TMRx_clk	
TMRx	
TMRx_postscaled	
TMRxIF	(1) (2) (1)
Note 1: 2:	Synchronization may take as many as 2 instruction cycles

23.6 Operation Examples

Unless otherwise specified, the following notes apply to the following timing diagrams:

- Both the prescaler and postscaler are set to 1:1 (both the CKPS and OUTPS bits in the TxCON register are cleared).
- The diagrams illustrate any clock except Fosc/4 and show clock-sync delays of at least two full cycles for both ON and Timer2_ers. When using Fosc/4, the clock-sync delay is at least one instruction period for Timer2_ers; ON applies in the next instruction period.
- The PWM Duty Cycle and PWM output are illustrated assuming that the timer is used for the PWM function of the CCP module as described in **Section 24.6** "**CCP/PWM Clock Selection**". The signals are not a part of the Timer2 module.

23.6.1 SOFTWARE GATE MODE

This mode corresponds to legacy Timer2 operation. The timer increments with each clock input when ON = 1 and does not increment when ON = 0. When the TMRx count equals the PRx period count the timer resets on the next clock and continues counting from 0. Operation with the ON bit software controlled is illustrated in Figure 23-4. With PRx = 5, the counter advances until TMRx = 5, and goes to zero with the next clock.



	Rev. 10-0001558 500/2014
MODE	060000
TMRx_clk	
Instruction ⁽¹⁾ -	(BSF) (BSF)
ON	
PRx	5
TMRx	0 (1)(2)(3)(4)(5)(0)(1)(2)(3)(4)(5)(0)(1)(2)(3)(4)(5)(0)(1)
TMRx_postscaled	
PWM Duty Cycle	3
PWM Output	
Note 1: set	BSF and BCF represent Bit-Set File and Bit-Clear File instructions executed by the CPU to or clear the ON bit of TxCON. CPU execution is asynchronous to the timer clock input.

23.6.2 HARDWARE GATE MODE

The Hardware Gate modes operate the same as the Software Gate mode except the TMRx_ers external signal can also gate the timer. When used with the CCP the gating extends the PWM period. If the timer is stopped when the PWM output is high then the duty cycle is also extended.

When MODE<4:0> = 00001 then the timer is stopped when the external signal is high. When MODE<4:0> = 00010 then the timer is stopped when the external signal is low.

Figure 23-5 illustrates the Hardware Gating mode for MODE<4:0> = 00001 in which a high input level starts the counter.



	5/30/2014
MODE	0600001
TMRx_clk	
TMRx_ers_	
PRx	5
TMRx	$0 \qquad 1 \\ 2 \\ 3 \\ 4 \\ 5 \\ 0 \\ 1 \\ 2 \\ 3 \\ 4 \\ 5 \\ 0 \\ 1 \\ 2 \\ 3 \\ 4 \\ 5 \\ 0 \\ 1 \\ 2 \\ 3 \\ 4 \\ 5 \\ 0 \\ 1 \\ 2 \\ 3 \\ 4 \\ 5 \\ 0 \\ 1 \\ 1 \\ 2 \\ 3 \\ 4 \\ 5 \\ 0 \\ 1 \\ 1 \\ 2 \\ 3 \\ 4 \\ 5 \\ 0 \\ 1 \\ 1 \\ 2 \\ 3 \\ 4 \\ 5 \\ 0 \\ 1 \\ 1 \\ 2 \\ 3 \\ 4 \\ 5 \\ 0 \\ 1 \\ 1 \\ 2 \\ 3 \\ 4 \\ 5 \\ 0 \\ 1 \\ 1 \\ 2 \\ 3 \\ 4 \\ 5 \\ 0 \\ 1 \\ 1 \\ 2 \\ 3 \\ 4 \\ 5 \\ 0 \\ 1 \\ 1 \\ 2 \\ 3 \\ 4 \\ 5 \\ 0 \\ 1 \\ 1 \\ 2 \\ 3 \\ 4 \\ 5 \\ 0 \\ 1 \\ 1 \\ 2 \\ 3 \\ 4 \\ 5 \\ 0 \\ 1 \\ 1 \\ 2 \\ 3 \\ 4 \\ 5 \\ 0 \\ 1 \\ 1 \\ 2 \\ 3 \\ 1 \\ 1 \\ 2 \\ 1 \\ 1 \\ 1 \\ 1 \\ 1 \\ 1 \\ 1$
TMRx_postscaled	
PWM Duty Cycle	3
PWM Output	

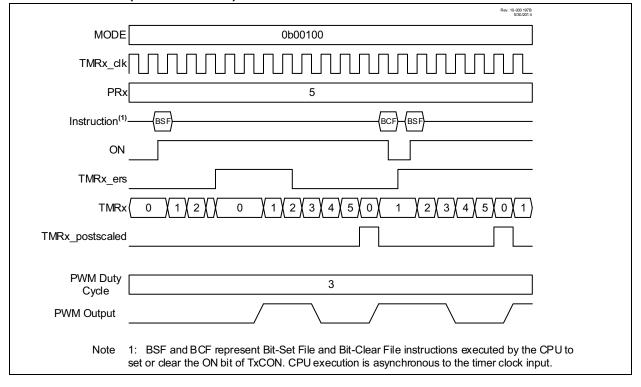
23.6.3 EDGE-TRIGGERED HARDWARE LIMIT MODE

In Hardware Limit mode the timer can be reset by the TMRx_ers external signal before the timer reaches the period count. Three types of Resets are possible:

- Reset on rising or falling edge (MODE<4:0>= 00011)
- Reset on rising edge (MODE<4:0> = 00100)
- Reset on falling edge (MODE<4:0> = 00101)

When the timer is used in conjunction with the CCP in PWM mode then an early Reset shortens the period and restarts the PWM pulse after a two clock delay. Refer to Figure 23-6.

FIGURE 23-6: EDGE-TRIGGERED HARDWARE LIMIT MODE TIMING DIAGRAM (MODE = 00100)

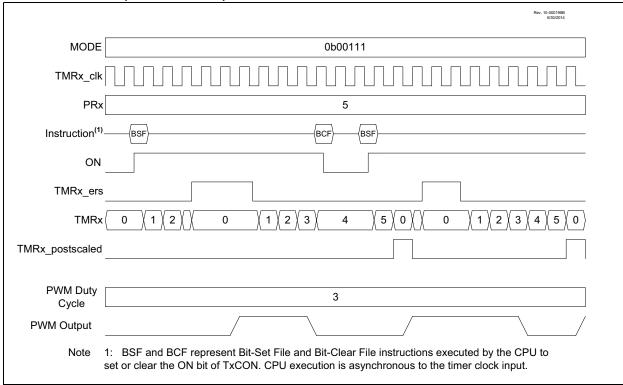


23.6.4 LEVEL-TRIGGERED HARDWARE LIMIT MODE

In the Level-Triggered Hardware Limit Timer modes the counter is reset by high or low levels of the external signal TMRx_ers, as shown in Figure 23-7. Selecting MODE<4:0> = 00110 will cause the timer to reset on a low level external signal. Selecting MODE<4:0> = 00111 will cause the timer to reset on a high level external signal. In the example, the counter is reset while TMRx_ers = 1. ON is controlled by BSF and BCF instructions. When ON = 0 the external signal is ignored.

When the CCP uses the timer as the PWM time base then the PWM output will be set high when the timer starts counting and then set low only when the timer count matches the CCPRx value. The timer is reset when either the timer count matches the PRx value or two clock periods after the external Reset signal goes true and stays true. The timer starts counting, and the PWM output is set high, on either the clock following the PRx match or two clocks after the external Reset signal relinquishes the Reset. The PWM output will remain high until the timer counts up to match the CCPRx pulse width value. If the external Reset signal goes true while the PWM output is high then the PWM output will remain high until the Reset signal is released allowing the timer to count up to match the CCPRx value.

FIGURE 23-7: LEVEL-TRIGGERED HARDWARE LIMIT MODE TIMING DIAGRAM (MODE = 00111)

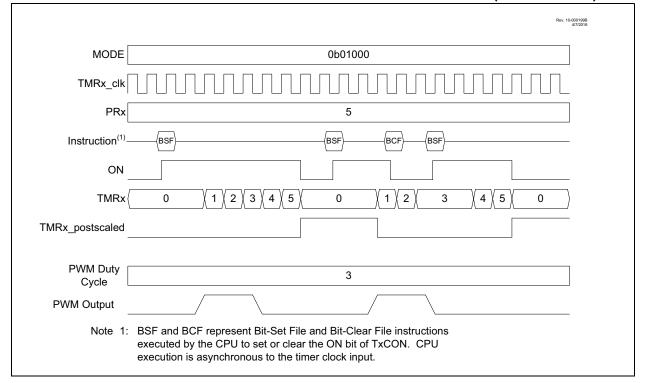


23.6.5 SOFTWARE START ONE-SHOT MODE

In One-Shot mode the timer resets and the ON bit is cleared when the timer value matches the PRx period value. The ON bit must be set by software to start another timer cycle. Setting MODE<4:0> = 01000 selects One-Shot mode which is illustrated in Figure 23-8. In the example, ON is controlled by BSF and BCF instructions. In the first case, a BSF instruction sets ON and the counter runs to completion and clears ON. In the second case, a BSF instruction starts the cycle, BCF/BSF instructions turn the counter off and on during the cycle, and then it runs to completion.

When One-Shot mode is used in conjunction with the CCP PWM operation the PWM pulse drive starts concurrent with setting the ON bit. Clearing the ON bit while the PWM drive is active will extend the PWM drive. The PWM drive will terminate when the timer value matches the CCPRx pulse width value. The PWM drive will remain off until software sets the ON bit to start another cycle. If software clears the ON bit after the CCPRx match but before the PRx match then the PWM drive will be extended by the length of time the ON bit remains cleared. Another timing cycle can only be initiated by setting the ON bit after it has been cleared by a PRx period count match.

FIGURE 23-8: SOFTWARE START ONE-SHOT MODE TIMING DIAGRAM (MODE = 01000)



23.6.6 EDGE-TRIGGERED ONE-SHOT MODE

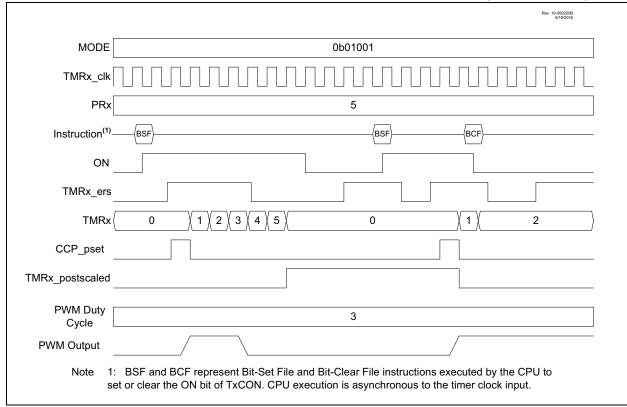
The Edge-Triggered One-Shot modes start the timer on an edge from the external signal input, after the ON bit is set, and clear the ON bit when the timer matches the PRx period value. The following edges will start the timer:

- Rising edge (MODE<4:0> = 01001)
- Falling edge (MODE<4:0> = 01010)
- Rising or Falling edge (MODE<4:0> = 01011)

If the timer is halted by clearing the ON bit then another TMRx_ers edge is required after the ON bit is set to resume counting. Figure 23-9 illustrates operation in the rising edge One-Shot mode.

When Edge-Triggered One-Shot mode is used in conjunction with the CCP then the edge-trigger will activate the PWM drive and the PWM drive will deactivate when the timer matches the CCPRx pulse width value and stay deactivated when the timer halts at the PRx period count match.

FIGURE 23-9: EDGE-TRIGGERED ONE-SHOT MODE TIMING DIAGRAM (MODE = 01001)



23.6.7 EDGE-TRIGGERED HARDWARE LIMIT ONE-SHOT MODE

In Edge-Triggered Hardware Limit One-Shot modes the timer starts on the first external signal edge after the ON bit is set and resets on all subsequent edges. Only the first edge after the ON bit is set is needed to start the timer. The counter will resume counting automatically two clocks after all subsequent external Reset edges. Edge triggers are as follows:

- Rising edge start and Reset (MODE<4:0> = 01100)
- Falling edge start and Reset (MODE<4:0> = 01101)

The timer resets and clears the ON bit when the timer value matches the PRx period value. External signal edges will have no effect until after software sets the ON bit. Figure 23-10 illustrates the rising edge hardware limit one-shot operation.

When this mode is used in conjunction with the CCP then the first starting edge trigger, and all subsequent Reset edges, will activate the PWM drive. The PWM drive will deactivate when the timer matches the CCPRx pulse-width value and stay deactivated until the timer halts at the PRx period match unless an external signal edge resets the timer before the match occurs.

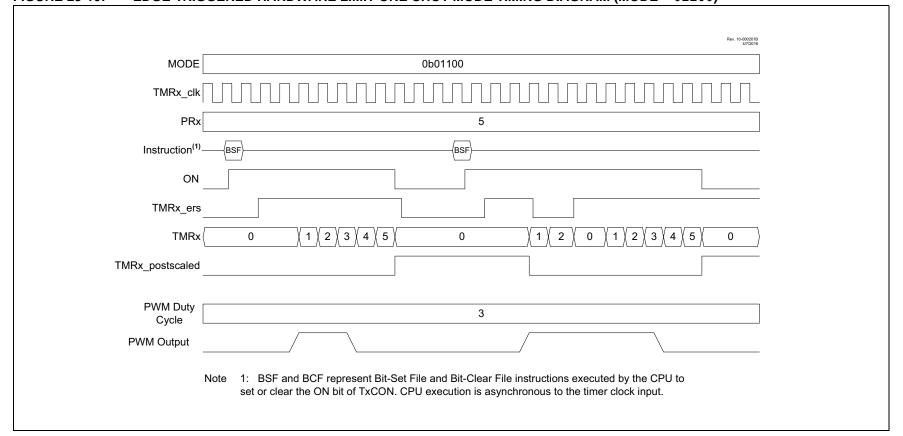


FIGURE 23-10: EDGE-TRIGGERED HARDWARE LIMIT ONE-SHOT MODE TIMING DIAGRAM (MODE = 01100)

23.6.8 LEVEL RESET, EDGE-TRIGGERED HARDWARE LIMIT ONE-SHOT MODES

In Level -Triggered One-Shot mode the timer count is reset on the external signal level and starts counting on the rising/falling edge of the transition from Reset level to the active level while the ON bit is set. Reset levels are selected as follows:

- Low Reset level (MODE<4:0> = 01110)
- High Reset level (MODE<4:0> = 01111)

When the timer count matches the PRx period count, the timer is reset and the ON bit is cleared. When the ON bit is cleared by either a PRx match or by software control a new external signal edge is required after the ON bit is set to start the counter.

When Level-Triggered Reset One-Shot mode is used in conjunction with the CCP PWM operation the PWM drive goes active with the external signal edge that starts the timer. The PWM drive goes inactive when the timer count equals the CCPRx pulse width count. The PWM drive does not go active when the timer count clears at the PRx period count match.

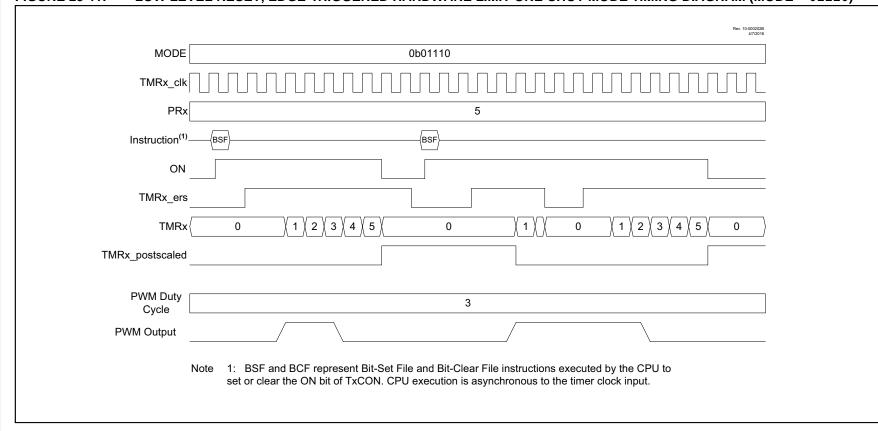


FIGURE 23-11: LOW LEVEL RESET, EDGE-TRIGGERED HARDWARE LIMIT ONE-SHOT MODE TIMING DIAGRAM (MODE = 01110)

PIC16(L)F1777/8/9

23.6.9 EDGE-TRIGGERED MONOSTABLE MODES

The Edge-Triggered Monostable modes start the timer on an edge from the external Reset signal input, after the ON bit is set, and stop incrementing the timer when the timer matches the PRx period value. The following edges will start the timer:

- Rising edge (MODE<4:0> = 10001)
- Falling edge (MODE<4:0> = 10010)
- Rising or Falling edge (MODE<4:0> = 10011)

When an Edge-Triggered Monostable mode is used in conjunction with the CCP PWM operation the PWM drive goes active with the external Reset signal edge that starts the timer, but will not go active when the timer matches the PRx value. While the timer is incrementing, additional edges on the external Reset signal will not affect the CCP PWM.

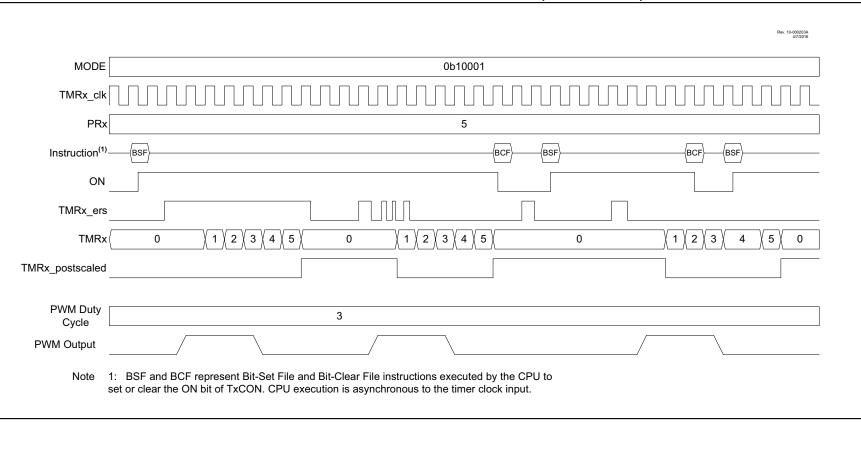


FIGURE 23-12: RISING EDGE-TRIGGERED MONOSTABLE MODE TIMING DIAGRAM (MODE = 10001)

PIC16(L)F1777/8/9

23.6.10 LEVEL-TRIGGERED HARDWARE LIMIT ONE-SHOT MODES

The Level-Triggered Hardware Limit One-Shot modes hold the timer in Reset on an external Reset level and start counting when both the ON bit is set and the external signal is not at the Reset level. If one of either the external signal is not in Reset or the ON bit is set then the other signal being set/made active will start the timer. Reset levels are selected as follows:

- Low Reset level (MODE<4:0> = 10110)
- High Reset level (MODE<4:0> = 10111)

When the timer count matches the PRx period count, the timer is reset and the ON bit is cleared. When the ON bit is cleared by either a PRx match or by software control the timer will stay in Reset until both the ON bit is set and the external signal is not at the Reset level.

When Level-Triggered Hardware Limit One-Shot modes are used in conjunction with the CCP PWM operation the PWM drive goes active with either the external signal edge or the setting of the ON bit, whichever of the two starts the timer.

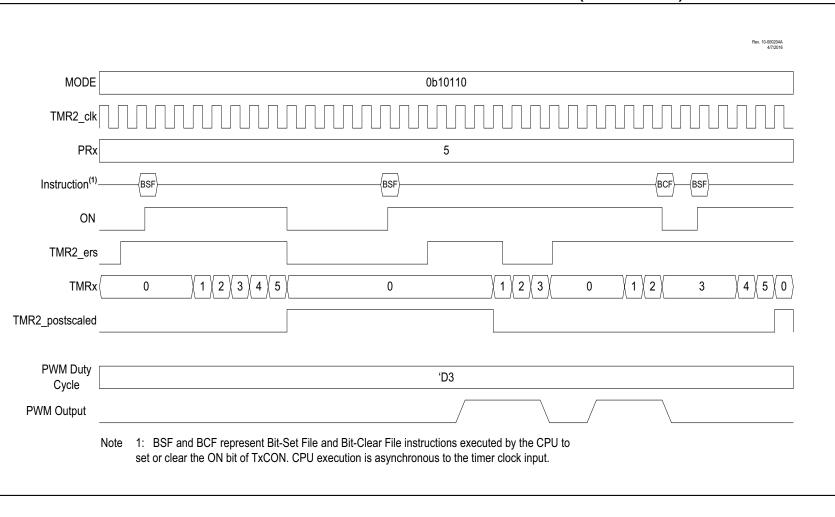


FIGURE 23-13: LEVEL-TRIGGERED HARDWARE LIMIT ONE-SHOT MODE TIMING DIAGRAM (MODE = 10110)

PIC16(L)F1777/8/9

23.7 PR2 Period Register

The PR2 period register (T2PR) is double-buffered. Software reads and writes the PR2 register. However, the timer uses a buffered PR2 register for operation. Software does not have direct access to the buffered PR2 register. The contents of the PR2 register are transferred to the buffer by any of the following events:

- A write to the TMR2 register
- A write to the TMR2CON register
- When TMR2 = PR2 buffer and the prescaler rolls over
- An external Reset event

23.8 Timer2 Operation During Sleep

When PSYNC = 1, Timer2 cannot be operated while the processor is in Sleep mode. The contents of the TMR2 and T2PR registers will remain unchanged while processor is in Sleep mode.

When PSYNC = 0, Timer2 will operate in Sleep as long as the clock source selected is also still running. Selecting the LFINTOSC, MFINTOSC, or HFINTOSC oscillator as the timer clock source will keep the selected oscillator running during Sleep.

23.9 Register Definitions: Timer2/4/6/8 Control

Long bit name prefixes for the Timer2/4/6/8 peripherals are shown in Table 23-2. Refer to **Section 1.1.2.2** "Long Bit Names" for more information

TABLE 23-2:

Peripheral	Bit Name Prefix
Timer2	T2
Timer4	T4
Timer6	Т6
Timer8	Т8

REGISTER 23-1: TxCLKCON: TIMERx CLOCK SELECTION REGISTER

U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
	—	—	—		CS<	3:0>	
bit 7	•		•	•			bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

- bit 7-4 Unimplemented: Read as '0'
- bit 3-0 CS<3:0>: Timerx Clock Selection bits

See Table 23-3.

TABLE 23-3: TIMERX CLOCK SOURCES

CS<3:0>	Timer2	Timer4	Timer6	Timer8
1100-1111	Reserved	Reserved	Reserved	Reserved
1011	LC4_out	LC4_out	LC4_out	LC4_out
1010	LC3_out	LC3_out	LC3_out	LC3_out
1001	LC2_out	LC2_out	LC2_out	LC2_out
1000	LC1_out	LC1_out	LC1_out	LC1_out
0111	ZCD_out	ZCD_out	ZCD_out	ZCD_out
0110	SOSC	SOSC	SOSC	SOSC
0101	MFINTOSC	MFINTOSC	MFINTOSC	MFINTOSC
0100	LFINTOSC	LFINTOSC	LFINTOSC	LFINTOSC
0011	HFINTOSC	HFINTOSC	HFINTOSC	HFINTOSC
0010	Fosc	Fosc	Fosc	Fosc
0001	Fosc/4	Fosc/4	Fosc/4	Fosc/4
0000	Pin selected by T2INPPS	Pin selected by T4INPPS	Pin selected by T6INPPS	Pin selected by T8INPPS

R/W/HC-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
0N ⁽¹⁾		CKPS<2:0>			OUTP	S<3:0>	
bit 7							bit 0
• • • • •							
Legend: R = Readable bit W = Writable bit					nanted bit rea	d oo 'O'	
R = Readable		x = Bit is unkr		-	nented bit, read		othor Doooto
u = Bit is unch	angeu					R/Value at all	other Resets
'1' = Bit is set		'0' = Bit is clea	ared	HC = Bit is cle	eared by hardv	vare	
bit 7	ON: Timerx	On hit					
	1 = Timerx						
		is off: all counte	rs and state n	nachines are res	set		
bit 6-4		: Timer2-type Cl					
	111 = 1:128	• •					
	110 = 1:64						
	101 = 1:32 I						
	100 = 1:16 F	Prescaler					
	011 = 1:8 P	rescaler					
	010 = 1:4 P	rescaler					
	001 = 1:2 P	rescaler					
	000 = 1:1 P	rescaler					
bit 3-0		>: Timerx Outpu	t Postscaler S	Select bits			
	1111 = 1 :16	Postscaler					
	1110 = 1:15						
	1101 = 1:14						
	1100 = 1:13						
	1011 = 1:12 1010 = 1:11						
	1010 - 1.11						
	1000 = 1:10 1000 = 1:9 F						
	0111 = 1:8 Postscaler						
	0110 = 1.7 Postscaler						
	0101 = 1:6 F	Postscaler					
	0100 = 1 :5 F	Postscaler					
	0011 = 1:4 F	Postscaler					
	0010 = 1:3 F						
	0001 = 1:2 F						
	0000 = 1:1 F	Postscaler					

REGISTER 23-2: TxCON: TIMERx CONTROL REGISTER

Note 1: In certain modes, the ON bit will be auto-cleared by hardware. See Section 23.6 "Operation Examples".

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
PSYNC ^{(1, 2}		CKSYNC ^(4, 5)	N/W-0/U		MODE<4:0> ^{(6, 7}		R/W-0/0
bit 7	, CKPUL	CKSTNC		1	MODE<4.020	,	hit 0
DIL 7							bit C
Legend:							
R = Readab	ole bit	W = Writable b	it	U = Unimpler	mented bit, read	1 as '0'	
u = Bit is un	changed	x = Bit is unkno	wn	-n/n = Value	at POR and BO	R/Value at all	other Resets
'1' = Bit is s	et	'0' = Bit is clear	ed				
dit /	bit 7 PSYNC: Timerx Prescaler Synchronization Enable bit ^(1, 2) 1 = TMRx Prescaler Output is synchronized to Fosc/4 0 = TMRx Prescaler Output is not synchronized to Fosc/4						
bit 6	 CKPOL: Timerx Clock Polarity Selection bit⁽³⁾ 1 = Falling edge of input clock clocks timer/prescaler 0 = Rising edge of input clock clocks timer/prescaler 						
bit 5							
bit 4-0 MODE<4:0>: Timerx Control Mode Selection bits ^(6, 7) See Table 23-1.							
Note 1:	Setting this bit e	ensures that reading	ng TMRx will	return a valid v	alue.		
2:	When this bit is '1', Timer2 cannot operate in Sleep mode.						
3:	3: CKPOL should not be changed while ON = 1.						
4:	Setting this bit e	ensures glitch-free	operation whether	nen the ON is e	nabled or disab	oled.	
5:	When this bit is a	set then the timer o	operation will	be delayed by t	wo TMRx input	clocks after the	e ON bit is set

REGISTER 23-3: TxHLT: TIMERx HARDWARE LIMIT CONTROL REGISTER

6: Unless otherwise indicated, all modes start upon ON = 1 and stop upon ON = 0 (stops occur without affect-

ing the value of TMRx).

7: When TMRx = PRx, the next clock clears TMRx, regardless of the operating mode.

REGISTER 23-4: TxRST: TIMERx EXTERNAL RESET SIGNAL SELECTION REGISTER

U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	—	—			RSEL<4:0>		
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-5 Unimplemented: Read as '0'

bit 4-0 **RSEL<4:0>:** TimerX External Reset Signal Source Selection bits See Table 23-4.

TABLE 23-4: EXTERNAL RESET SOURCES

RSEL<4:0>	Timer2	Timer4	Timer6	Timer8
11111	Reserved	Reserved	Reserved	Reserved
11110	Reserved	Reserved	Reserved	Reserved
11101	LC4_out	LC4_out	LC4_out	LC4_out
11100	LC3_out	LC3_out	LC3_out	LC3_out
11011	LC2_out	LC2_out	LC2_out	LC2_out
11010	LC1_out	LC1_out	LC1_out	LC1_out
11001	ZCD_out	ZCD_out	ZCD_out	ZCD_out
11000 (1)	sync_C8OUT	sync_C8OUT	sync_C8OUT	sync_C8OUT
10111 (1)	sync_C7OUT	sync_C7OUT	sync_C7OUT	sync_C7OUT
10110	sync_C6OUT	sync_C6OUT	sync_C6OUT	sync_C6OUT
10101	sync_C5OUT	sync_C5OUT	sync_C5OUT	sync_C5OUT
10100	sync_C4OUT	sync_C4OUT	sync_C4OUT	sync_C4OUT
10011	sync_C3OUT	sync_C3OUT	sync_C3OUT	sync_C3OUT
10010	sync_C2OUT	sync_C2OUT	sync_C2OUT	sync_C2OUT
10001	sync_C1OUT	sync_C1OUT	sync_C1OUT	sync_C1OUT
10000 (1)	PWM12_out	PWM12_out	PWM12_out	PWM12_out
01111	PWM11_out	PWM11_out	PWM11_out	PWM11_out
01110	PWM6_out	PWM6_out	PWM6_out	PWM6_out
01101	PWM5_out	PWM5_out	PWM5_out	PWM5_out
01100 (1)	PWM10_out	PWM10_out	PWM10_out	PWM10_out
01011	PWM9_out	PWM9_out	PWM9_out	PWM9_out
01010	PWM4_out	PWM4_out	PWM4_out	PWM4_out
01001	PWM3_out	PWM3_out	PWM3_out	PWM3_out
01000(1)	CCP8_out	CCP8_out	CCP8_out	CCP8_out
00111	CCP7_out	CCP7_out	CCP7_out	CCP7_out
00110	CCP2_out	CCP2_out	CCP2_out	CCP2_out
00101	CCP1_out	CCP1_out	CCP1_out	CCP1_out
00100	TMR8_postscaled	TMR8_postscaled	TMR8_postscaled	Reserved
00011	TMR6_postscaled	TMR6_postscaled	Reserved	TMR6_postscaled
00010	TMR4_postscaled	Reserved	TMR4_postscaled	TMR4_postscaled
00001	Reserved	TMR2_postscaled	TMR2_postscaled	TMR2_postscaled
00000	Pin selected byT2INPPS	Pin selected by T4INPPS	Pin selected by T6INPPS	Pin selected by T6INPPS

Note 1: PIC16LF1777/9 only.

PIC16(L)F1777/8/9

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
CCP1CON	EN	—	OUT	FMT		MODE	=<3:0>		319
CCP2CON	EN	—	OUT	FMT		MODE	=<3:0>		319
CCP7CON	EN	—	OUT	FMT		MODE	=<3:0>		319
CCP8CON ⁽¹⁾	EN	—	OUT	FMT		MODE	=<3:0>		319
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	132
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	CCP1IE	TMR2IE	TMR1IE	133
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	CCP1IF	TMR2IF	TMR1IF	139
T2PR	Timer2 Modu	le Period Reg	jister						287*
TMR2	Holding Regi	ister for the 8-l	oit TMR2 Regi	ster					287*
T2CON	ON		CKPS<2:0>			OUTP	S<3:0>		307
T2CLKCON	—	—	_	_		CS<	:3:0>		306
T2RST	—	_				RSEL<4:0>			309
T2HLT	PSYNC	CKPOL	CKSYNC			308			
T4PR	Timer4 Modu	le Period Reg	ister						287*
TMR4	Holding Regi	ister for the 8-I	oit TMR4 Regi	ster					287*
T4CON	ON		CKPS<2:0>		OUTPS<3:0>				307
T4CLKCON	—	—	_	_		CS<	:3:0>		306
T4RST	—	—	_				309		
T4HLT	PSYNC	CKPOL	CKSYNC			308			
T6PR	Timer6 Modu	le Period Reg	ister						287*
TMR6	Holding Regi	ister for the 8-I	oit TMR6 Regi	ster					287*
T6CON	ON		CKPS<2:0>			OUTP	S<3:0>		307
T6CLKCON	—	—	_	_		CS<	:3:0>		306
T6RST	—	—	_			RSEL<4:0>			309
T6HLT	PSYNC	CKPOL	CKSYNC		MODE<4:0>				308
T8PR	Timer6 Modu	le Period Reg	ister						287*
TMR8	Holding Regi	ister for the 8-l	oit TMR6 Regi	ster					287*
T8CON	ON		CKPS<2:0>			OUTP	S<3:0>		307
T8CLKCON	—	—	—	—		CS<	:3:0>		306
T8RST	_	_	_			RSEL<4:0>			309
T8HLT	PSYNC	CKPOL	CKSYNC			MODE<4:0>			308

TABLE 23-5: SUMMARY OF REGISTERS ASSOCIATED WITH TIMER2

Legend: — = unimplemented location, read as '0'. Shaded cells are not used for Timer2 module.

* Page provides register information.

Note 1: PIC16LF1777/9 only.

24.0 CAPTURE/COMPARE/PWM MODULES

The Capture/Compare/PWM module is a peripheral which allows the user to time and control different events, and to generate Pulse-Width Modulation (PWM) signals. In Capture mode, the peripheral allows the timing of the duration of an event. The Compare mode allows the user to trigger an external event when a predetermined amount of time has expired. The PWM mode can generate Pulse-Width Modulated signals of varying frequency and duty cycle.

Device	CCP1	CCP2	CCP7	CCP8		
PIC16(L)F1778	•	•	٠			
PIC16(L)F1777/9	•	•	٠	•		

TABLE 24-1: AVAILABLE CCP MODULES

- Note 1: In devices with more than one CCP module, it is very important to pay close attention to the register names used. A number placed after the module acronym is used to distinguish between separate modules. For example, the CCP1CON and CCP2CON control the same operational aspects of two completely different CCP modules.
 - 2: Throughout this section, generic references to a CCP module in any of its operating modes may be interpreted as being equally applicable to a CCPx module. Register names, module signals, I/O pins, and bit names may use the generic designator 'x' to indicate the use of a numeral to distinguish a particular module, when required.

24.1 Capture Mode

The Capture mode function described in this section is available and identical for all CCP modules.

Capture mode makes use of the 16-bit Timer1 resource. When an event occurs on the CCPx input, the 16-bit CCPRxH:CCPRxL register pair captures and stores the 16-bit value of the TMR1H:TMR1L register pair, respectively. An event is defined as one of the following and is configured by the MODE<3:0> bits of the CCPxCON register:

- Every edge (rising or falling)
- Every falling edge
- Every rising edge
- Every 4th rising edge
- Every 16th rising edge

The CCPx capture input signal is configured by the CTS bits of the CCPxCAP register with the following options:

- CCPx pin
- Comparator 1 output (C1_OUT_sync)
- Comparator 2 output (C2_OUT_sync)
- Comparator 7 output (C7_OUT_sync)
- Comparator 8 output (C8_OUT_sync) (PIC16(L)F1777/9 only)
- LC2_output
- LC3_output
- Interrupt-on-change interrupt trigger (IOC_interrupt)

When a capture is made, the Interrupt Request Flag bit CCPxIF of the PIRx register is set. The interrupt flag must be cleared in software. If another capture occurs before the value in the CCPRxH:CCPRxL register pair is read, the old captured value is overwritten by the new captured value.

Figure 24-1 shows a simplified diagram of the capture operation.

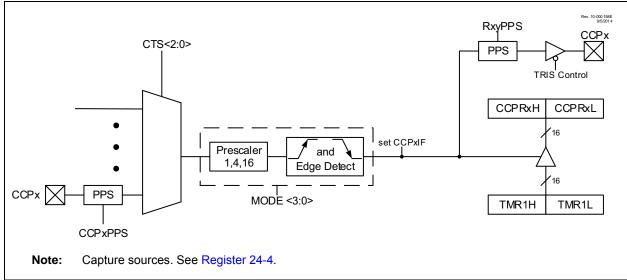
24.1.1 CCP PIN CONFIGURATION

In Capture mode, select the interrupt source using the CTS bits of the CCPxCAP register. If the CCPx pin is chosen, it should be configured as an input by setting the associated TRIS control bit.

Note: If the CCPx pin is configured as an output, a write to the port can cause a capture condition.

PIC16(L)F1777/8/9

FIGURE 24-1: CAPTURE MODE OPERATION BLOCK DIAGRAM



24.1.2 TIMER1 MODE RESOURCE

Timer1 must be running in Timer mode or Synchronized Counter mode for the CCP module to use the capture feature. In Asynchronous Counter mode, the capture operation may not work.

See Section 22.0 "Timer1/3/5 Module with Gate Control" for more information on configuring Timer1.

24.1.3 SOFTWARE INTERRUPT MODE

When the Capture mode is changed, a false capture interrupt may be generated. The user should keep the CCPxIE interrupt enable bit of the PIEx register clear to avoid false interrupts. Additionally, the user should clear the CCPxIF interrupt flag bit of the PIRx register following any change in operating mode.

Note: Clocking Timer1 from the system clock (Fosc) should not be used in Capture mode. In order for Capture mode to recognize the trigger event on the CCPx pin, Timer1 must be clocked from the instruction clock (Fosc/4) or from an external clock source.

24.1.4 CCP PRESCALER

There are four prescaler settings specified by the MODE<3:0> bits of the CCPxCON register. Whenever the CCP module is turned off, or the CCP module is not in Capture mode, the prescaler counter is cleared. Any Reset will clear the prescaler counter.

Switching from one capture prescaler to another does not clear the prescaler and may generate a false interrupt. To avoid this unexpected operation, turn the module off by clearing the EN bit of the CCPxCON register before changing the prescaler.

24.1.5 CAPTURE DURING SLEEP

Capture mode depends upon the Timer1 module for proper operation. There are two options for driving the Timer1 module in Capture mode. It can be driven by the instruction clock (Fosc/4), or by an external clock source.

When Timer1 is clocked by Fosc/4, Timer1 will not increment during Sleep. When the device wakes from Sleep, Timer1 will continue from its previous state.

Capture mode will operate during Sleep when Timer1 is clocked by an external clock source.

24.1.6 ALTERNATE PIN LOCATIONS

This module incorporates I/O pins that can be moved to other locations with the use of the PPS controls. See **Section 12.0 "Peripheral Pin Select (PPS) Module"** for more details.

24.1.7 CAPTURE OUTPUT

Whenever a capture occurs, the output of the CCP will go high for a period equal to one system clock period (1/Fosc). This output is available as an input signal to the following peripherals:

- ADC Trigger
- COG
- PRG
- DSM
- CLC
- · Op Amp override
- Timer2/4/6/8 Reset
- Any device pins

In addition, the CCP output can be output to any pin with that pin's PPS control.

24.2 Compare Mode

The Compare mode function described in this section is available and identical for all CCP modules.

Compare mode makes use of the 16-bit Timer1 resource. The 16-bit value of the CCPRxH:CCPRxL register pair is constantly compared against the 16-bit value of the TMR1H:TMR1L register pair. When a match occurs, one of the following events can occur:

- · Toggle the CCPx output
- Set the CCPx output
- Clear the CCPx output
- · Pulse the CCPx output
- Generate a Software Interrupt
- Auto-conversion Trigger

The action on the pin is based on the value of the MODE<3:0> control bits of the CCPxCON register. At the same time, the interrupt flag CCPxIF bit is set.

All Compare modes can generate an interrupt.

Figure 24-2 shows a simplified diagram of the compare operation.

24.2.1 AUTO-CONVERSION TRIGGER

When Auto-Conversion Trigger mode is chosen (CCPxM<3:0> = 1011), the CCPx module does the following:

- Resets Timer1
- Starts an ADC conversion if ADC is enabled

The CCPx module does not assert control of the CCPx pin in this mode.

The Auto-conversion Trigger output of the CCP occurs immediately upon a match between the TMR1H, TMR1L register pair and the CCPRxH:CCPRxL register pair. The TMR1H:TMR1L register pair is not reset until the next rising edge of the Timer1 clock. The Auto-conversion Trigger output starts an ADC conversion (if the ADC module is enabled). This allows the CCPRxH:CCPRxL register pair to effectively provide a 16-bit programmable period register for Timer1.

Refer to Section 16.2.5 "Auto-Conversion Trigger" for more information.

- Note 1: The Auto-conversion Trigger from the CCP module does not set interrupt flag bit TMR1IF of the PIR1 register.
 - 2: Removing the match condition by changing the contents of the CCPRxH and CCPRxL register pair, between the clock edge that generates the Auto-conversion Trigger and the clock edge that generates the Timer1 Reset, will preclude the Reset from occurring.

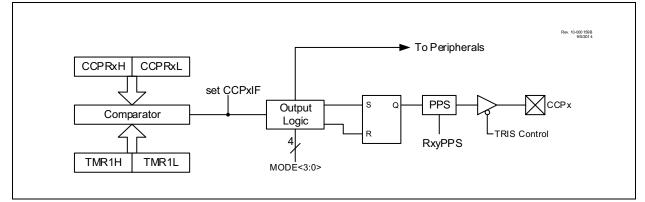
24.2.2 CCPx PIN CONFIGURATION

The user must configure the CCPx pin as an output by clearing the associated TRIS bit.

The CCPx pin function can be moved to alternate pins using the PPS controls. See Section 12.0 "Peripheral Pin Select (PPS) Module" for more detail.

Note: Clearing the CCPxCON register will force the CCPx compare output latch to the default low level. This is not the PORT I/O data latch.

FIGURE 24-2: COMPARE MODE OPERATION BLOCK DIAGRAM



24.2.3 TIMER1 MODE RESOURCE

In Compare mode, Timer1 must be running in either Timer mode or Synchronized Counter mode. The compare operation may not work in Asynchronous Counter mode.

See Section 22.0 "Timer1/3/5 Module with Gate Control" for more information on configuring Timer1.

Note: Clocking Timer1 from the system clock (Fosc) should not be used in Compare mode. In order for Compare mode to recognize the trigger event on the CCPx pin, TImer1 must be clocked from the instruction clock (Fosc/4) or from an external clock source.

24.2.4 SOFTWARE INTERRUPT MODE

When Generate Software Interrupt mode is chosen (MODE<3:0> = 1010), the CCPx module does not assert control of the CCPx pin (see the CCPxCON register).

24.2.5 COMPARE DURING SLEEP

The Compare mode is dependent upon the system clock (Fosc) for proper operation. Since Fosc is shut down during Sleep mode, the Compare mode will not function properly during Sleep.

24.2.6 ALTERNATE PIN LOCATIONS

This module incorporates I/O pins that can be moved to other locations with the use of the PPS controls. See **Section 12.0 "Peripheral Pin Select (PPS) Module"** for more detail.

24.2.7 CAPTURE OUTPUT

When in Compare mode, the CCP will provide an output upon the 16-bit value of the CCPRxH:CCPRxL register pair, matching the TMR1H:TMR1L register pair. The compare output depends on which Compare mode the CCP is configured as. If the MODE bits of CCPxCON register are equal to '1011' or '1010', the CCP module will output high, while TMR1 is equal to the CCPRxH:CCPRxL register pair. This means that the pulse width is determined by the TMR1 prescaler. If the MODE bits of CCPxCON are equal to '0001' or '0010', the output will toggle upon a match, going from '0' to '1' or vice-versa. If the MODE bits of CCPxCON are equal to '1001', the output is cleared on a match, and if the MODE bits are equal to '1000', the output is set on a match. This output is available to the following peripherals:

- ADC Trigger
- COG
- PRG
- DSM
- CLC
- Op Amp override
- Timer2/4/6/8 Reset
- · Any device pins

24.3 PWM Overview

Pulse-Width Modulation (PWM) is a scheme that provides power to a load by switching quickly between fully on and fully off states. The PWM signal resembles a square wave where the high portion of the signal is considered the on state and the low portion of the signal is considered the off state. The high portion, also known as the pulse width, can vary in time and is defined in steps. A larger number of steps applied, which lengthens the pulse width, also supplies more power to the load. Lowering the number of steps applied, which shortens the pulse width, supplies less power. The PWM period is defined as the duration of one complete cycle or the total amount of on and off time combined. PWM resolution defines the maximum number of steps that can be present in a single PWM period. A higher resolution allows for more precise control of the pulse width time and in turn the power that is applied to the load.

The term duty cycle describes the proportion of the on time to the off time and is expressed in percentages, where 0% is fully off and 100% is fully on. A lower duty cycle corresponds to less power applied and a higher duty cycle corresponds to more power applied.

Figure 24-3 shows a typical waveform of the PWM signal.

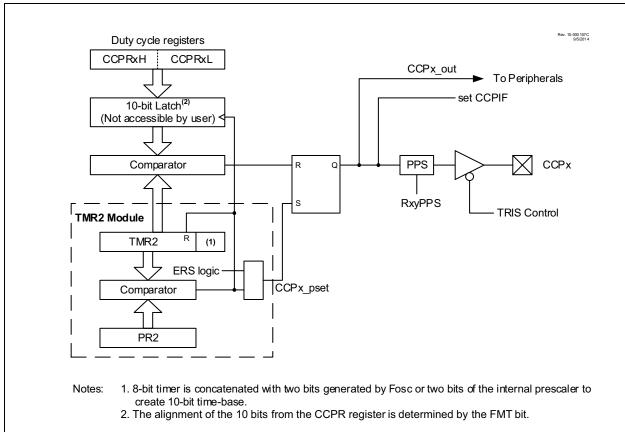


FIGURE 24-3: SIMPLIFIED PWM BLOCK DIAGRAM

24.3.1 STANDARD PWM OPERATION

The standard PWM function described in this section is available and identical for all CCP modules.

The standard PWM mode generates a Pulse-Width Modulation (PWM) signal on the CCPx pin with up to ten bits of resolution. The period, duty cycle, and resolution are controlled by the following registers:

- T2PR/T4PR/T6PR/T8PR registers
- T2CON/T4CON/T6CON/T8CON registers
- CCPRxH:CCPRxL register pair

Figure 24-3 shows a simplified block diagram of PWM operation.

Note 1: The corresponding TRIS bit must be cleared to enable the PWM output on the CCPx pin.

2: Clearing the CCPxCON register will relinquish control of the CCPx pin.

24.3.2 SETUP FOR PWM OPERATION

The following steps should be taken when configuring the CCP module for standard PWM operation:

- 1. Disable the CCPx pin output driver by setting the associated TRIS bit.
- 2. Select the timer associated with the PWM by setting the CCPTMRS register.
- 3. Load the associated T2PR/T4PR/T6PR/T8PR register with the PWM period value.
- Configure the CCP module for the PWM mode by loading the CCPxCON register with the appropriate values.
- 5. Load the CCPRxH:CCPRxL register pair with the PWM duty cycle value.
- 6. Configure and start the timer selected in step 2:
 - Clear the timer interrupt flag bit of the PIRx register. See Note below.
 - Configure the CKPS bits of the TxCON register with the Timer prescale value.
 - Enable the Timer by setting the ON bit of the TxCON register.
- 7. Enable PWM output pin:
 - Wait until the Timer overflows and the timer interrupt bit of the PIRx register is set. See Note below.
 - Enable the CCPx pin output driver by clearing the associated TRIS bit.
 - **Note:** In order to send a complete duty cycle and period on the first PWM output, the above steps must be included in the setup sequence. If it is not critical to start with a complete PWM signal on the first output, then step 6 may be ignored.

24.4 CCP/PWM Clock Selection

The PIC16(L)F1777/8/9 allows each individual CCP and PWM module to select the timer source that controls the module. Each module has an independent selection.

As there are up to four 8-bit timers with auto-reload (Timer2/4/6/8). The PWM mode on the CCP and PWM modules can use any of these timers.

The CCPTMRS register is used to select which timer is used.

24.4.1 USING THE TMR2/4/6/8 WITH THE CCP MODULE

This device has a new version of the TMR2 module that has many new modes, which allow for greater customization and control of the PWM signals than older parts. Refer to **Section 23.6** "**Operation Examples**" for examples of PWM signal generation using the different modes of Timer2. The CCP operation requires that the timer used as the PWM time base has the Fosc/4 clock source selected.

24.4.2 PWM PERIOD

The PWM period is specified by the T2PR/T4PR/T6PR/T8PR register of Timer2/4/6/8. The PWM period can be calculated using the formula of Equation 24-1.

EQUATION 24-1: PWM PERIOD

$$PWM Period = [(PR2) + 1] \bullet 4 \bullet Tosc \bullet$$
$$(TMR2 Prescale Value)$$

Note 1: Tosc = 1/Fosc

When TMR2/4/6/8 is equal to its respective T2PR/T4PR/T6PR/T8PR register, the following three events occur on the next increment cycle:

- TMR2/4/6/8 is cleared
- The CCPx pin is set. (Exception: If the PWM duty cycle = 0%, the pin will not be set.)
- The PWM duty cycle is latched from the CCPRxH:CCPRxL pair into the internal 10-bit latch.

Note:	The Timer postscaler (see Figure 24-1) is
	not used in the determination of the PWM
	frequency.

24.4.3 PWM DUTY CYCLE

The PWM duty cycle is specified by writing a 10-bit value to two registers: the CCPRxH:CCPRxL register pair. Where the particular bits go is determined by the FMT bit of the CCPxCON register. If FMT = 0, the two Most Significant bits of the duty cycle value should be written to bits <1:0> of the CCPRxH register and the remaining eight bits to the CCPRxL register. If FMT = 1, the Least Significant two bits of the duty cycle should be written to bits <7:6> of the CCPRxL register and the Most Significant eight bits to the CCPRxH register. This is illustrated in Figure 24-4. These bits can be written at any time. The duty cycle value is not latched into the internal latch until after the period completes (i.e., a match between T2PR/T4PR/T6PR/T8PR and TMR2/4/6/8 registers occurs).

Equation 24-2 is used to calculate the PWM pulse width. Equation 24-3 is used to calculate the PWM duty cycle ratio.

EQUATION 24-2: PULSE WIDTH

Pulse Width = CCPRxH:CCPRxL • Tosc

• (TMR2 Prescale Value)

EQUATION 24-3: DUTY CYCLE RATIO

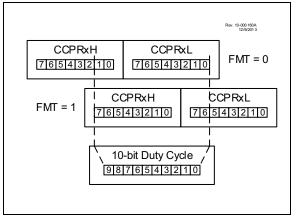
Duty Cycle Ratio = $\frac{(CCPRxH:CCPRxL)}{4(PRx+1)}$

The PWM duty cycle registers are double buffered for glitchless PWM operation.

The 8-bit timer TMR2/4/6/8 register is concatenated with either the 2-bit internal system clock (FOSC), or two bits of the prescaler, to create the 10-bit time base. The system clock is used if the Timer2/4/6/8 prescaler is set to 1:1.

When the 10-bit time base matches the internal buffer register, then the CCPx pin is cleared (see Figure 24-3).

FIGURE 24-4: CCPx DUTY CYCLE ALIGNMENT



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24.4.4 PWM RESOLUTION

The resolution determines the number of available duty cycles for a given period. For example, a 10-bit resolution will result in 1024 discrete duty cycles, whereas an 8-bit resolution will result in 256 discrete duty cycles.

The maximum PWM resolution is ten bits when T2PR/T4PR/T6PR/T8PR is 255. The resolution is a function of the T2PR/T4PR/T6PR/T8PR register value as shown by Equation 24-4.

EQUATION 24-4: PWM RESOLUTION

Resolution =
$$\frac{\log[4(PR2 + 1)]}{\log(2)}$$
 bits

Note: If the pulse-width value is greater than the period, the assigned PWM pin(s) will remain unchanged.

TABLE 24-2: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (Fosc = 20 MHz)

PWM Frequency	1.22 kHz	4.88 kHz	19.53 kHz	78.12 kHz	156.3 kHz	208.3 kHz
Timer Prescale	16	4	1	1	1	1
T2PR Value	0xFF	0xFF	0xFF	0x3F	0x1F	0x17
Maximum Resolution (bits)	10	10	10	8	7	6

TABLE 24-3: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (Fosc = 8 MHz)

PWM Frequency	1.22 kHz	4.90 kHz	19.61 kHz	76.92 kHz	153.85 kHz	200.0 kHz
Timer Prescale	16	4	1	1	1	1
T2PR Value	0x65	0x65	0x65	0x19	0x0C	0x09
Maximum Resolution (bits)	8	8	8	6	5	5

24.4.5 CHANGES IN SYSTEM CLOCK FREQUENCY

The PWM frequency is derived from the system clock frequency. Any changes in the system clock frequency will result in changes to the PWM frequency. See Section 5.0 "Oscillator Module (with Fail-Safe Clock Monitor)" for additional details.

24.4.6 EFFECTS OF RESET

Any Reset will force all ports to Input mode and the CCP registers to their Reset states.

24.4.7 PWM OUTPUT

The output of the CCP in PWM mode is the PWM signal generated by the module and described above. This output is available to the following peripherals:

- ADC Trigger
- COG
- PRG
- DSM
- CLC
- Op Amp override
- Timer2/4/6/8 Reset
- · Any device pins

24.5 Register Definitions: CCP Control

R/W-0/0	U-0	R-x	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
EN		OUT	FMT		MODE	=<3:0>	
bit 7							bit
Legend:							
R = Readable	e bit	W = Writable	bit	U = Unimpler	nented bit, read	1 as '0'	
u = Bit is uncl		x = Bit is unk		•	at POR and BC		other Reset
'1' = Bit is set	•	'0' = Bit is cle					
bit 7	EN: CCPx M	Iodule Enable	bit				
	1 = CCPx is	enabled					
	0 = CCPx is	disabled					
bit 6	Unimpleme	nted: Read as	'0'				
bit 5	OUT: CCPx	Output Data bi	t (read-only)				
bit 4	FMT: CCPW	(Pulse-Width)	Alignment bit				
	<u>If MODE = P</u>	WM Mode					
				the MSB of the the LSB of the l			
bit 3-0	MODE<3:0>	: CCPx Mode	Selection bits				
	11xx = PW	M mode					
		npare mode: P npare mode: P					
				compare match. mpare match. C			
	1000 - 001	inpute mode. St					the mode.
		ture mode: eve					
		oture mode: eventure		dge			
		oture mode: eve					
	0100 000						
		oture mode: eve					
		npare mode: to					
				nd clear TMR1 c ts CCPx module		hackwards co	mnatibility
	0000 - Oa p	and compare/					mpanomy)

REGISTER 24-1: CCPxCON: CCPx CONTROL REGISTER

REGISTER 24-2: CCPRxL: CCPx LOW BYTE REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0			
			CCP	R<7:0>						
bit 7							bit 0			
Legend:										
R = Readable	e bit	W = Writable	bit	U = Unimpler	nented bit, read	d as '0'				
u = Bit is unchanged		x = Bit is unkr	x = Bit is unknown		-n/n = Value at POR and BOR/Value at all other Reset					
'1' = Bit is set	t	'0' = Bit is clea	ared							
bit 7-0	<u>MODE = Ca</u>	<u>pture Mode</u>								
	CCPRxL<7:	0>: LSB of capt	ured TMR1 v	alue						
	MODE = Co	mpare Mode								
	CCPRxL<7:0>: LSB compared to TMR1 value									
	MODE = PWM Mode && FMT = 0									
	CCPRxL<7:	0>: CCPW<7:0	> – Pulse wid	th Least Signific	ant eight bits					
	MODE = PWM Mode && FMT = 1									

CCPRxL<7:6>: CCPW<1:0> - Pulse width Least Significant two bits

CCPRxL<5:0>: Not used

REGISTER 24-3: CCPRxH: CCPx HIGH BYTE REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
			CCPR	<15:8>			
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Reset
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 MODE = Capture Mode CCPRxH<7:0>: MSB of captured TMR1 value MODE = Compare Mode CCPRxH<7:0>: MSB compared to TMR1 value MODE = PWM Mode && FMT = 0 CCPRxH<7:2>: Not used CCPRxH<7:2>: Not used CCPRxH<7:0>: CCPW<9:8> – Pulse width Most Significant two bits MODE = PWM Mode && FMT = 1 CCPRxH<7:0>: CCPW<9:2> – Pulse width Most Significant eight bits

U-0 U-0 U-0 U-0 R/W-0/0 R/W-0/0 R/W-0/0 R/W-0/0 _ CTS<3:0> ____ ____ ____ bit 7 bit 0 Legend: R = Readable bit W = Writable bit U = Unimplemented bit, read as '0' u = Bit is unchanged -n/n = Value at POR and BOR/Value at all other Reset x = Bit is unknown '1' = Bit is set '0' = Bit is cleared bit 7-4 Unimplemented: Read as '0' bit 3-0 CTS<3:0>: Capture Trigger Input Selection bits 1101 = IOC event 1100 = LC4 output 1011 = LC3_output 1010 = LC2_output 1001 = LC1_output 1000 = C8_sync_out⁽¹⁾ 0111 = C7_sync_out⁽¹⁾ 0110 = C6_sync_out 0101 = C5_sync_out 0100 = C4_sync_out 0011 = C3_sync_out

REGISTER 24-4: CCPxCAP: CCPx CAPTURE INPUT SELECTION REGISTER

Note 1: PIC16LF1777/9 only.

0010 = C2_sync_out 0001 = C1 sync out

0000 = Pin selected with the CCPxPPS register

24.6 CCP/PWM Clock Selection

This device allows each individual CCP and PWM module to select the timer source that controls the module. Each module has an independent selection.

As there are four 8-bit timers with auto-reload (Timer2, Timer4, Timer6 and Timer8). The PWM mode on the CCP and 10-bit PWM modules can use any of these timers.

The CCPTMRS register is used to select which timer is used.

24.7 Register Definitions: CCP/PWM Timers Control

REGISTER 24-5: CCPTMRS1: PWM TIMER SELECTION CONTROL REGISTER 1

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
C8TSEL	<1:0>(1)	C7TSE	L<1:0>	C2TSE	EL<1:0>	C1TSE	:L<1:0>
bit 7							bit 0

Legend:			
R = Read	able bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is u	unchanged	-n/n = Value at POR and BOR/Value at all other Resets	
'1' = Bit is	set	'0' = Bit is cleared	
bit 7-6	11 = CCP8 10 = CCP8 01 = CCP8	0>: CCP8 (PWM8) Timer S B is based off Timer8 in PWI B is based off Timer6 in PWI B is based off Timer4 in PWI	M mode M mode M mode
bit 5-4	C7TSEL<1: 11 = CCP7	B is based off Timer2 in PWN 0>: CCP7 (PWM7) Timer S 7 is based off Timer8 in PWN bis based off Timer8 in PWN	election bits M mode
	01 = CCPT $00 = CCPT$	7 is based off Timer6 in PWI 7 is based off Timer4 in PWI 7 is based off Timer2 in PWI	M mode M mode
bit 3-2	11 = CCP2 10 = CCP2 01 = CCP2	0>: CCP2 (PWM2) Timer S 2 is based off Timer8 in PWI 2 is based off Timer6 in PWI 2 is based off Timer4 in PWI 2 is based off Timer2 in PWI	M mode M mode M mode
bit 1-0	11 = CCP 10 = CCP 01 = CCP	0>: CCP1 (PWM1) Timer S I is based off Timer8 in PWI I is based off Timer6 in PWI I is based off Timer4 in PWI I is based off Timer2 in PWI	M mode M mode M mode
Note 1:	PIC16(L)F1777/	9 only.	

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0		
P10TSEL<1:0> ⁽¹⁾		P9TSEL<1:0>		P4TSEL<1:0>		P3TSEL<1:0>			
bit 7							bit 0		
Legend:									
R = Readable bit		W = Writable bit		U = Unimplemented bit, read as '0'					
u = Bit is unchanged		x = Bit is unknown -n/n = Value at POR and				R/Value at all	other Resets		
'1' = Bit is set		'0' = Bit is clea	ared						
bit 7-6	P10TSEL<1:	0>: PWM10 Tir	mer Selection I	oits ⁽¹⁾					
		0 is based off T							
	10 = PWM10 is based off Timer6 in PWM mode								
	01 = PWM10 is based off Timer4 in PWM mode								
	00 = PWM10 is based off Timer2 in PWM mode								
bit 5-4	P9TSEL<1:0	>: PWM9 Time	r Selection bits	6					
		is based off Til							
		is based off Til							
		is based off Til							
		is based off Til							
bit 3-2		>: PWM4 Time		-					
		is based off Til							
		is based off Til							
		is based off Til							
		is based off Ti							
bit 1-0		>: PWM3 Time		-					
		is based off Til							
		is based off Til							
		is based off Til							
	00 = PVVIVI3	is based off Ti		node					

REGISTER 24-6: CCPTMRS2: PWM TIMER SELECTION CONTROL REGISTER 2

Note 1: PIC16(L)F1777/9 only.

PIC16(L)F1777/8/9

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page		
_	_	_	_		321					
EN	_	OUT FMT MODE<3:0>								
Capture/Compare/PWM Register x (LSB)										
Capture/Compare/PWM Register x (MSB)										
C8TSEL	.<1:0> ⁽¹⁾	C7TSEL<1:0>		C2TSEL<1:0>		C1TSEL<1:0>		323		
P10TSEL<1:0>(1)		P9TSEL<1:0>		P4TSEL<1:0>		P3TSEL<1:0>		323		
GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	132		
TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	CCP1IE	TMR2IE	TMR1IE	133		
OSFIE	C2IE	C1IE	COG1IE	BCL1IE	C4IE	C3IE	CCP2IE	134		
CCP8IE ⁽¹⁾	CCP7IE	COG4IE ⁽¹⁾	COG3IE	C8IE ⁽¹⁾	C7IE ⁽¹⁾	C6IE	C5IE	137		
TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	CCP1IF	TMR2IF	TMR1IF	139		
OSFIF	C2IF	C1IF	COG1IF	BCL1IF	C4IF	C3IF	CCP2IF	140		
CCP8IF ⁽¹⁾	CCP7IF	COG4IF ⁽¹⁾	COG3IF	C8IF ⁽¹⁾	C7IF ⁽¹⁾	C6IF	C5IF	143		
Timer2 Period Register										
ON CKPS<2:0>					307					
Timer2 Module Register								287		
Timer4 Period Register										
ON CKPS<2:0> OUTPS<3:0>								307		
Timer4 Module Register										
Timer6 Period Register										
ON		CKPS<2:0>			307					
Timer6 Modu	Timer6 Module Register									
Timer8 Period Register										
ON	CKPS<2:0> OUTPS<3:0>							307		
Timer8 Module Register										
	EN Capture/Com Capture/Com Capture/Com Capture/Com C8TSEL P10TSEI GIE TMR1GIE OSFIE CCP8IE ⁽¹⁾ TMR1GIF OSFIF CCP8IF ⁽¹⁾ Timer2 Perio ON Timer2 Perio ON Timer4 Perio ON Timer6 Perio ON	Image: Arrow of the second	Image <t< td=""><td>Image: constraint of the sector of the se</td><td>$\begin{array}{ c c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c } \hline \end{tabular} \\ \hline tabula$</td><td>$\begin{array}{ c c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c } \hline \hline \begin{tabular}{ c c$</td><td>$\begin{array}{ c c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c } \hline \hline \begin{tabular}{ c c } \hline \hline \begin{tabular}{ c c } \hline \hline \begin{tabular}{ c c c } \hline \hline \begin{tabular}{ c c c } \hline \hline \begin{tabular}{ c c c } \hline \hline \begin{tabular}{ c c } \hline \hline \begin{tabular}{ c c c } \hline \hline \begin{tabular}{ c c c } \hline \hline \begin{tabular}{ c c c } \hline \hline \begin{tabular}{ c c c } \hline \hline \begi$</td><td>CTS<3:0>ENOUTFMTMODE<3:0>Capture/Compare/PWM Register x (LSB)Capture/Compare/PWM Register x (MSB)Cattset/compare/PWM Register x (MSB)Capture/Compare/PWM Register x (MSB)Cattset/compare/PWM Register x (MSB)Cattset/compare/PWM Register x (MSB)Capture/Compare/PWM Register x (MSB)Cattset/compare/PWM Register X (MSB)GifePEIETMROIFINTEIOCIETMROIFIMR1GIEADIETMROIETXIESSP1IECCP1IETMR1IEOSFIFC2IECOG4IE(1)COG3IECale(1)CCP1IFTMR2IFTMR1IFOSFIFC2IFCOG4IE(1)COG3IFCale(1)CAIFC3IFCSIFTimer2 PeriodCKPS<2:0>OUTPS<3:0>Timer4CoUTPS<3:0>Timer6ONCKPS<2:0>OUTPS<3:0>OUTPS<3:0>Timer6ONCKPS<2:0>OUTPS<3:0>Timer6CaleseCaleseONCKPS<2:0>OUTPS<3:0>OUTPS<3:0>Timer6 MedisterCipset/compare<t< td=""></t<></td></t<>	Image: constraint of the sector of the se	$ \begin{array}{ c c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c } \hline \end{tabular} \\ \hline tabula$	$\begin{array}{ c c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c } \hline \hline \begin{tabular}{ c c $	$ \begin{array}{ c c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c c } \hline \begin{tabular}{ c c } \hline \hline \begin{tabular}{ c c } \hline \hline \begin{tabular}{ c c } \hline \hline \begin{tabular}{ c c c } \hline \hline \begin{tabular}{ c c c } \hline \hline \begin{tabular}{ c c c } \hline \hline \begin{tabular}{ c c } \hline \hline \begin{tabular}{ c c c } \hline \hline \begin{tabular}{ c c c } \hline \hline \begin{tabular}{ c c c } \hline \hline \begin{tabular}{ c c c } \hline \hline \begi$	CTS<3:0>ENOUTFMTMODE<3:0>Capture/Compare/PWM Register x (LSB)Capture/Compare/PWM Register x (MSB)Cattset/compare/PWM Register x (MSB)Capture/Compare/PWM Register x (MSB)Cattset/compare/PWM Register x (MSB)Cattset/compare/PWM Register x (MSB)Capture/Compare/PWM Register x (MSB)Cattset/compare/PWM Register X (MSB)GifePEIETMROIFINTEIOCIETMROIFIMR1GIEADIETMROIETXIESSP1IECCP1IETMR1IEOSFIFC2IECOG4IE(1)COG3IECale(1)CCP1IFTMR2IFTMR1IFOSFIFC2IFCOG4IE(1)COG3IFCale(1)CAIFC3IFCSIFTimer2 PeriodCKPS<2:0>OUTPS<3:0>Timer4CoUTPS<3:0>Timer6ONCKPS<2:0>OUTPS<3:0>OUTPS<3:0>Timer6ONCKPS<2:0>OUTPS<3:0>Timer6CaleseCaleseONCKPS<2:0>OUTPS<3:0>OUTPS<3:0>Timer6 MedisterCipset/compare <t< td=""></t<>		

TABLE 24-4: SUMMARY OF REGISTERS ASSOCIATED WITH STANDARD PWM

Legend: — = Unimplemented location, read as '0'. Shaded cells are not used by the PWM.

* Page provides register information.

Note 1: PIC16(L)F1777/9 only.

25.0 10-BIT PULSE-WIDTH MODULATION (PWM) MODULE

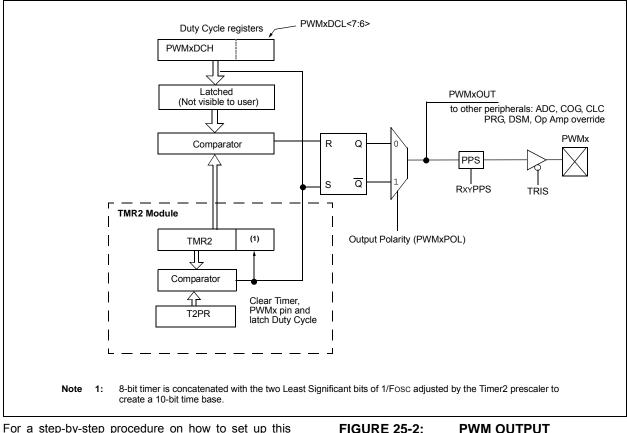
The 10-bit PWM module generates a Pulse-Width Modulated signal determined by the duty cycle, period, and resolution that are configured by the following registers:

- T2PR
- T2CON
- PWMxDCH
- PWMxDCL
- PWMxCON

Figure 25-1 shows a simplified block diagram of PWM operation.

Figure 25-2 shows a typical waveform of the PWM signal.





For a step-by-step procedure on how to set up this module for PWM operation, refer to Section 25.1.9 "Setup for PWM Operation using PWMx Output Pins".

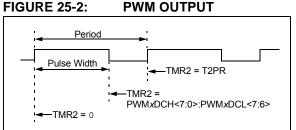


TABLE 25-1: AVAILABLE 10-BIT PWM MODULES

Device	PWM3	PWM4	PWM9	PWM10
PIC16(L)F1778	•	•	•	
PIC16(L)F1777/9	•	•	•	•

25.1 PWMx Pin Configuration

All PWM outputs are multiplexed with the PORT data latch. The user must configure the pins as outputs by clearing the associated TRIS bits.

25.1.1 FUNDAMENTAL OPERATION

The PWM module produces a 10-bit resolution output. Timer2 and T2PR set the period of the PWM. The PWMxDCL and PWMxDCH registers configure the duty cycle. The period is common to all PWM modules, whereas the duty cycle is independently controlled.

Note: The Timer2 postscaler is not used in the determination of the PWM frequency. The postscaler could be used to have a servo update rate at a different frequency than the PWM output.

All PWM outputs associated with Timer2 are set when TMR2 is cleared. Each PWMx is cleared when TMR2 is equal to the value specified in the corresponding PWMxDCH (8 MSb) and PWMxDCL<7:6> (2 LSb) registers. When the value is greater than or equal to T2PR, the PWM output is never cleared (100% duty cycle).

Note: The PWMxDCH and PWMxDCL registers are double buffered. The buffers are updated when Timer2 matches T2PR. Care should be taken to update both registers before the timer match occurs.

25.1.2 PWM OUTPUT POLARITY

The output polarity is inverted by setting the PWMxPOL bit of the PWMxCON register.

25.1.3 PWM PERIOD

The PWM period is specified by the T2PR register of Timer2. The PWM period can be calculated using the formula of Equation 25-1.

EQUATION 25-1: PWM PERIOD

 $PWM Period = [T2PR + 1] \bullet 4 \bullet Tosc \bullet$

(TMR2 Prescale Value)

Note: Tosc = 1/Fosc

When TMR2 is equal to T2PR, the following three events occur on the next increment cycle:

- TMR2 is cleared
- The PWM output is active. (Exception: When the PWM duty cycle = 0%, the PWM output will remain inactive.)
- The PWMxDCH and PWMxDCL register values are latched into the buffers.

Note:	The Timer2 postscaler has no effect on the
	PWM operation.

25.1.4 PWM DUTY CYCLE

The PWM duty cycle is specified by writing a 10-bit value to the PWMxDCH and PWMxDCL register pair. The PWMxDCH register contains the eight MSbs and the PWMxDCL<7:6>, the two LSbs. The PWMxDCH and PWMxDCL registers can be written to at any time.

Equation 25-2 is used to calculate the PWM pulse width.

Equation 25-3 is used to calculate the PWM duty cycle ratio.

EQUATION 25-2: PULSE WIDTH

 $Pulse Width = (PWMxDCH:PWMxDCL<7:6>) \bullet$

TOSC • (TMR2 Prescale Value)

Note: Tosc = 1/Fosc

EQUATION 25-3: DUTY CYCLE RATIO

$$Duty Cycle Ratio = \frac{(PWMxDCH:PWMxDCL<7:6>)}{4(T2PR+1)}$$

The 8-bit timer TMR2 register is concatenated with the two Least Significant bits of 1/Fosc, adjusted by the Timer2 prescaler to create the 10-bit time base. The system clock is used if the Timer2 prescaler is set to 1:1.

25.1.5 PWM RESOLUTION

The resolution determines the number of available duty cycles for a given period. For example, a 10-bit resolution will result in 1024 discrete duty cycles, whereas an 8-bit resolution will result in 256 discrete duty cycles.

The maximum PWM resolution is ten bits when T2PR is 255. The resolution is a function of the T2PR register value as shown by Equation 25-4.

EQUATION 25-4: PWM RESOLUTION

Resolution = $\frac{\log[4(T2PR+1)]}{\log(2)}$ bits

Note: If the pulse-width value is greater than the period the assigned PWM pin(s) will remain unchanged.

PWM Frequency	0.31 kHz	4.88 kHz	19.53 kHz	78.12 kHz	156.3 kHz	208.3 kHz
Timer Prescale	64	4	1	1	1	1
T2PR Value	0xFF	0xFF	0xFF	0x3F	0x1F	0x17
Maximum Resolution (bits)	10	10	10	8	7	6.6

TABLE 25-3: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (Fosc = 8 MHz)

PWM Frequency	0.31 kHz	4.90 kHz	19.61 kHz	76.92 kHz	153.85 kHz	200.0 kHz
Timer Prescale	64	4	1	1	1	1
T2PR Value	0x65	0x65	0x65	0x19	0x0C	0x09
Maximum Resolution (bits)	8	8	8	6	5	5

25.1.6 OPERATION IN SLEEP MODE

In Sleep mode, the TMR2 register will not increment and the state of the module will not change. If the PWMx pin is driving a value, it will continue to drive that value. When the device wakes up, TMR2 will continue from its previous state.

25.1.7 CHANGES IN SYSTEM CLOCK FREQUENCY

The PWM frequency is derived from the system clock frequency (Fosc). Any changes in the system clock frequency will result in changes to the PWM frequency. Refer to Section 5.0 "Oscillator Module (with Fail-Safe Clock Monitor)" for additional details.

25.1.8 EFFECTS OF RESET

Any Reset will force all ports to Input mode and the PWM registers to their Reset states.

25.1.9 SETUP FOR PWM OPERATION USING PWMx OUTPUT PINS

The following steps should be taken when configuring the module for PWM operation using the PWMx output pins:

- 1. Disable the PWMx pin output driver(s) by setting the associated TRIS bit(s).
- 2. Clear the PWMxCON register.
- 3. Load the T2PR register with the PWM period value.
- Load the PWMxDCH register and bits <7:6> of the PWMxDCL register with the PWM duty cycle value.
- 5. Configure and start Timer2:
- Clear the TMR2IF interrupt flag bit of the PIR1
 register. See Note below.
- Configure the CKPS bits of the T2CON register with the Timer2 prescale value.
- Enable Timer2 by setting the ON bit of the T2CON register.
- Enable PWM output pin and wait until Timer2 overflows, TMR2IF bit of the PIR1 register is set. See Note below.
- Enable the PWMx pin output driver(s) by clearing the associated TRIS bit(s) and setting the desired pin PPS control bits.
- 8. Configure the PWM module by loading the PWMxCON register with the appropriate values.
 - **Note 1:** In order to send a complete duty cycle and period on the first PWM output, the above steps must be followed in the order given. If it is not critical to start with a complete PWM signal, then move Step 8 to replace Step 4.
 - **2:** For operation with other peripherals only, disable PWMx pin outputs.

25.1.10 SETUP FOR PWM OPERATION TO OTHER DEVICE PERIPHERALS

The following steps should be taken when configuring the module for PWM operation to be used by other device peripherals:

- 1. Disable the PWMx pin output driver(s) by setting the associated TRIS bit(s).
- 2. Clear the PWMxCON register.
- 3. Load the T2PR register with the PWM period value.
- Load the PWMxDCH register and bits <7:6> of the PWMxDCL register with the PWM duty cycle value.
- 5. Configure and start Timer2:
 - Clear the TMR2IF interrupt flag bit of the PIR1 register. See Note below.
 - Configure the CKPS bits of the T2CON register with the Timer2 prescale value.
 - Enable Timer2 by setting the ON bit of the T2CON register.
- 6. Enable PWM output pin:
- Wait until Timer2 overflows, TMR2IF bit of the PIR1 register is set. See Note below.
- 7. Configure the PWM module by loading the PWMxCON register with the appropriate values.

Note: In order to send a complete duty cycle and period on the first PWM output, the above steps must be included in the setup sequence. If it is not critical to start with a complete PWM signal on the first output, then step 6 may be ignored.

25.2 Register Definitions: 10-Bit PWM Control

Long bit name prefixes for the DSM peripherals are shown in Table 25-4. Refer to **Section 1.1.2.2** "Long **Bit Names**" for more information

TABLE 25-4:

Peripheral	Bit Name Prefix
PWM3	PWM3
PWM4	PWM4
PWM9	PWM9
PWM10 ⁽¹⁾	PWM10

Note 1: PIC16(L)F1777/9 only.

REGISTER 25-1: PWMxCON: PWM CONTROL REGISTER

R/W-0/0	U-0	R-0/0	R/W-0/0	U-0	U-0	U-0	U-0
EN	—	OUT	POL	—	—	—	—
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7	EN: PWM Module Enable bit

1 = PWM module is enabled0 = PWM module is disabled

bit 6 Unimplemented: Read as '0'

bit 5 **OUT:** PWM module output level when bit is read.

- bit 4 **POL:** PWMx Output Polarity Select bit
 - 1 = PWM output is active-low
 - 0 = PWM output is active-high
- bit 3-0 Unimplemented: Read as '0'

REGISTER 25-2: PWMxDCH: PWM DUTY CYCLE HIGH BITS

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
			DC	<9:2>			
bit 7							bit 0
Legend:							
R = Readable	bit	W = Writable b	bit	U = Unimplen	nented bit, read	l as '0'	
u = Bit is unch	nanged	x = Bit is unkn	own	-n/n = Value a	at POR and BO	R/Value at all o	other Resets
'1' = Bit is set		'0' = Bit is clea	ired				

bit 7-0 DC<9:2>: PWM Duty Cycle Most Significant bits These bits are the MSbs of the PWM duty cycle. The two LSbs are found in the PWMxDCL Register.

REGISTER 25-3: PWMxDCL: PWM DUTY CYCLE LOW BITS

R/W-x/u	R/W-x/u	U-0	U-0	U-0	U-0	U-0	U-0
DC<	1:0>	—	—	—	—	—	—
bit 7				bit 0			
Legend-							

Legenu.		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-6 DC<1:0>: PWM Duty Cycle Least Significant bits

These bits are the LSbs of the PWM duty cycle. The MSbs are found in the PWMxDCH Register.

bit 5-0 Unimplemented: Read as '0'

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
CCPTMRS2	P10TSE	L<1:0> ⁽¹⁾	P9TSE	L<1:0>	P4TSE	EL<1:0>	P3TSE	:L<1:0>	323
PWM3CON	EN	—	OUT	POL	—	_	—	_	329
PWM3DCH				DC<	9:2>				330
PWM3DCL	DC<	:1:0>	—	_	—	—	—	_	330
PWM4CON	EN	—	OUT	POL	_	—	_		329
PWM4DCH				DC<	:9:2>				330
PWM4DCL	DC<	:1:0>	—		_	—	_		330
PWM9CON	EN	—	OUT	POL	_	_	_	_	329
PWM9DCH		DC<9:2>							
PWM9DCL	DC<	:1:0>	—	_	_	_	_	_	330
PWM10CON ⁽¹⁾	EN	—	OUT	POL	_	_	_	_	329
PWM10DCH ⁽¹⁾				DC<	9:2>				330
PWM10DCL ⁽¹⁾	DC<	:1:0>	_		_	_	_	_	330
RxyPPS	_	—			RxyPP	205			
TxCON	ON		CKPS<2:0>		OUTPS<3:0>				307
TxCLKCON	_	—	—	_	CS<3:0>				306
TxPR	TMRx Period Register							287	
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	176
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	181
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	186

TABLE 25-5: SUMMARY OF REGISTERS ASSOCIATED WITH 10-BIT PWM

Legend: – = Unimplemented locations, read as '0', u = unchanged, x = unknown. Shaded cells are not used by the PWM. **Note 1:** PIC16LF1777/9 only.

26.0 16-BIT PULSE-WIDTH MODULATION (PWM) MODULE

The Pulse-Width Modulation (PWM) module generates a pulse-width modulated signal determined by the phase, duty cycle, period, and offset event counts that are contained in the following registers:

- PWMxPH register
- PWMxDC register
- PWMxPR register
- PWMxOF register

Figure 26-1 shows a simplified block diagram of the PWM operation. Each PWM module has four modes of operation:

- Standard
- Set On Match
- Toggle On Match
- · Center Aligned

TABLE 26-1:AVAILABLE 16-BIT PWM
MODULES

Device	PWM5	PWM6	PWM11	PWM12
PIC16(L)F1778	•	•	•	
PIC16(L)F1777/9	•	•	٠	•

FIGURE 26-1: 16-BIT PWM BLOCK DIAGRAM

For a more detailed description of each PWM mode, refer to **Section 26.2 "PWM Modes"**.

Each PWM module has four offset modes:

- · Independent Run
- Slave Run with Synchronous Start
- One-Shot Slave with Synchronous Start
- Continuous Run Slave with Synchronous Start and Timer Reset

Using the offset modes, each PWM module can offset its waveform relative to any other PWM module in the same device. For a more detailed description of the offset modes refer to **Section 26.3 "Offset Modes"**.

Every PWM module has a configurable reload operation to ensure all event count buffers change at the end of a period, thereby avoiding signal glitches. Figure 26-2 shows a simplified block diagram of the reload operation. For a more detailed description of the reload operation, refer to Section **Section 26.4 "Reload Operation**".

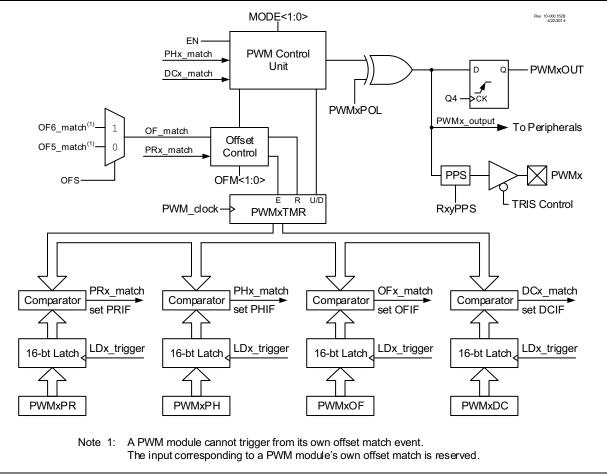
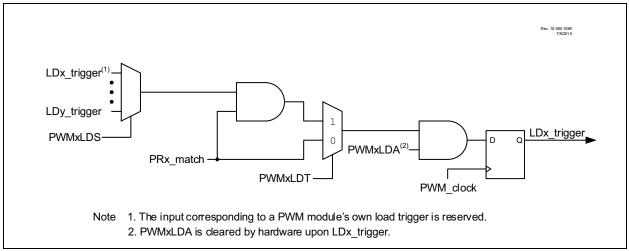


FIGURE 26-2: LOAD TRIGGER BLOCK DIAGRAM



26.1 Fundamental Operation

The PWM module produces a 16-bit resolution pulse-width modulated output.

Each PWM module has an independent timer driven by a selection of clock sources determined by the PWMxCLKCON register (Register 26-4). The timer value is compared to event count registers to generate the various events of a the PWM waveform, such as the period and duty cycle. For a block diagram describing the clock sources refer to Figure 26-3.

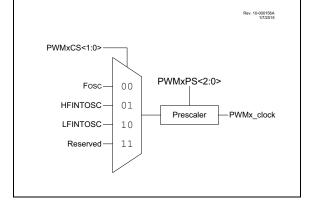
Each PWM module can be enabled individually using the EN bit of the PWMxCON register, or several PWM modules can be enabled simultaneously using the MPWMxEN bits of the PWMEN register.

The current state of the PWM output can be read using the OUT bit of the PWMxCON register. In some modes this bit can be set and cleared by software giving additional software control over the PWM waveform. This bit is synchronized to Fosc/4 and therefore does not change in real time with respect to the PWM_clock.

Note: If PWM_clock > Fosc/4, the OUT bit may not accurately represent the output state of the PWM.

FIGURE 26-3:

PWM CLOCK SOURCE BLOCK DIAGRAM



26.1.1 PWMx PIN CONFIGURATION

This device uses the PPS control circuitry to route peripherals to any device I/O pin. Select the desired pin, or pins, for PWM output with the device pin RxyPPS control registers (Register 12-2).

All PWM outputs are multiplexed with the PORT data latch, so the pins must also be configured as outputs by clearing the associated PORT TRIS bits.

The slew rate feature may be configured to optimize the rate to be used in conjunction with the PWM outputs. High-speed output switching is attained by clearing the associated PORT SLRCON bits.

The PWM outputs can be configured to be open-drain outputs by setting the associated PORT ODCON bits.

26.1.2 PWMx Output Polarity

The output polarity is inverted by setting the POL bit of the PWMxCON register. The polarity control affects the PWM output even when the module is not enabled.

26.2 PWM Modes

PWM modes are selected with MODE<1:0> bits of the PWMxCON register (Register 26-1).

In all PWM modes an offset match event can also be used to synchronize the PWMxTMR in three offset modes. See **Section 26.3 "Offset Modes**" for more information.

26.2.1 STANDARD MODE

The Standard mode (MODE = 00) selects a single phase PWM output. The PWM output in this mode is determined by when the period, duty cycle, and phase counts match the PWMxTMR value. The start of the duty cycle occurs on the phase match and the end of the duty cycle occurs on the duty cycle match. The period match resets the timer. The offset match can also be used to synchronize the PWMxTMR in the offset modes. See Section 26.3 "Offset Modes" for more information.

Equation 26-1 is used to calculate the PWM period in Standard mode.

Equation 26-2 is used to calculate the PWM duty cycle ratio in Standard mode.

EQUATION 26-1: PWM PERIOD IN STANDARD MODE

$$Period = \frac{(PWMxPR + 1) \cdot Prescale}{PWM_clock}$$

EQUATION 26-2: PWM DUTY CYCLE IN STANDARD MODE

$$Duty Cycle = \frac{(PWMxDC - PWMxPH)}{PWMxPR + 1}$$

A detailed timing diagram for Standard mode is shown in Figure 26-4.

26.2.2 SET ON MATCH MODE

The Set On Match mode (MODE = 01) generates an active output when the phase count matches the PWMxTMR value. The output stays active until the OUT bit of the PWMxCON register is cleared or the PWM module is disabled. The duty cycle count has no effect in this mode. The period count only determines the maximum PWMxTMR value above which no phase matches can occur.

The PWMxOUT bit can be used to set or clear the output of the PWM in this mode. Writes to this bit will take place on the next rising edge of the PWM_clock after the bit is written.

A detailed timing diagram for Set On Match is shown in Figure 26-5.

26.2.3 TOGGLE ON MATCH MODE

The Toggle On Match mode (MODE = 10) generates a 50% duty cycle PWM with a period twice as long as that computed for the standard PWM mode. Duty cycle count has no effect in this mode. The phase count determines how many PWMxTMR periods after a period event the output will toggle.

Writes to the OUT bit of the PWMxCON register will have no effect in this mode.

A detailed timing diagram for Toggle On Match is shown in Figure 26-6.

26.2.4 CENTER ALIGNED MODE

The Center Aligned mode (MODE = 11) generates a PWM waveform that is centered in the period. In this mode the period is two times the PWMxPR count. The PWMxTMR counts up to the period value then counts back down to 0. The duty cycle count determines both the start and end of the active PWM output. The start of the duty cycle occurs at the match event when PWMxTMR is incrementing and the duty cycle ends at the match event when PWMxTMR is decrementing. The incrementing match value is the period count minus the duty cycle count. The decrementing match value is the incrementing match value plus 1.

Equation 26-3 is used to calculate the PWM period in Center Aligned mode.

EQUATION 26-3: PWM PERIOD IN CENTER ALIGNED MODE

$$Period = \frac{(PWMxPR + 1) \cdot 2 \cdot Prescale}{PWM_clock}$$

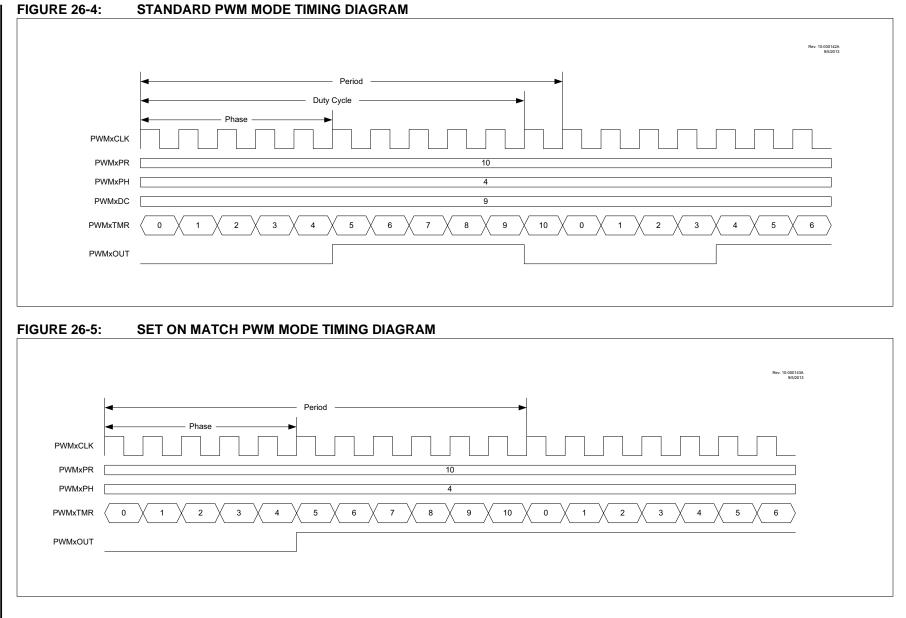
Equation 26-4 is used to calculate the PWM duty cycle ratio in Center Aligned mode

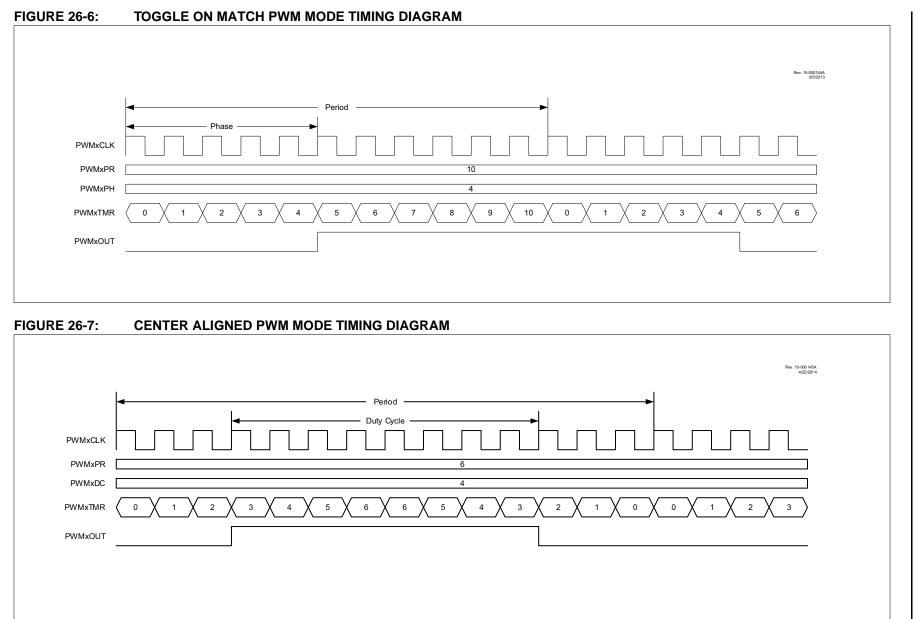
EQUATION 26-4: PWM DUTY CYCLE IN CENTER ALIGNED MODE

$$Duty Cycle = \frac{PWMxDC \cdot 2}{(PWMxPR + 1) \cdot 2}$$

Writes to PWMxOUT will have no effect in this mode.

A detailed timing diagram for Center Aligned mode is shown in Figure 26-7.





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26.3 Offset Modes

The offset modes provide the means to adjust the waveform of a slave PWM module relative to the waveform of a master PWM module in the same device.

26.3.1 INDEPENDENT RUN MODE

In Independent Run mode (OFM = 00), the PWM module is unaffected by the other PWM modules in the device. The PWMxTMR associated with the PWM module in this mode starts counting as soon as the EN bit associated with this PWM module is set and continues counting until the EN bit is cleared. Period events reset the PWMxTMR to zero after which the timer continues to count.

A detailed timing diagram of this mode used with Standard PWM mode is shown in Figure 26-8.

26.3.2 SLAVE RUN MODE WITH SYNC START

In Slave Run mode with Sync Start (OFM = 01), the slave PWMxTMR waits for the master's OF_match event. When this event occurs, if the EN bit is set, the PWMxTMR begins counting and continues to count until software clears the EN bit. Slave period events reset the PWMxTMR to zero after which the timer continues to count.

A detailed timing diagram of this mode used with Standard PWM mode is shown in Figure 26-9.

26.3.3 ONE-SHOT SLAVE MODE WITH SYNC START

In One-Shot Slave mode with Synchronous Start (OFM = 10), the slave PWMxTMR waits until the master's OF_match event. The timer then begins counting, starting from the value that is already in the timer, and continues to count until the period match event. When the period event occurs the timer resets to zero and stops counting. The timer then waits until the next master OF_match event after which it begins counting again to repeat the cycle.

A detailed timing diagram of this mode used with Standard PWM mode is shown in Figure 26-10.

26.3.4 CONTINUOUS RUN SLAVE MODE WITH SYNC START AND TIMER RESET

In Continuous Run Slave mode with Synchronous Start and Timer Reset (OFM = 11) the slave PWMxTMR is inhibited from counting after the slave PWM enable is set. The first master OF_match event starts the slave PWMxTMR. Subsequent master OF_match events reset the slave PWMxTMR timer value back to 1 after which the slave PWMxTMR continues to count. The next master OF_match event resets the slave PWMxTMR back to 1 to repeat the cycle. Slave period events that occur before the master's OF_match event will reset the slave PWMxTMR to zero after which the timer will continue to count. Slaves operating in this mode must have a PWMxPH register pair value equal to or greater than 1, otherwise the phase match event will not occur precluding the start of the PWM output duty cycle.

The offset timing will persist if both the master and slave PWMxPR values are the same and the Slave Offset mode is changed to Independent Run mode while the PWM module is operating.

A detailed timing diagram of this mode used in Standard PWM mode is shown in Figure 26-11.

Note:	Unexpected results will occur if the slave
	PWM_clock is a higher frequency than the
	master PWM_clock.

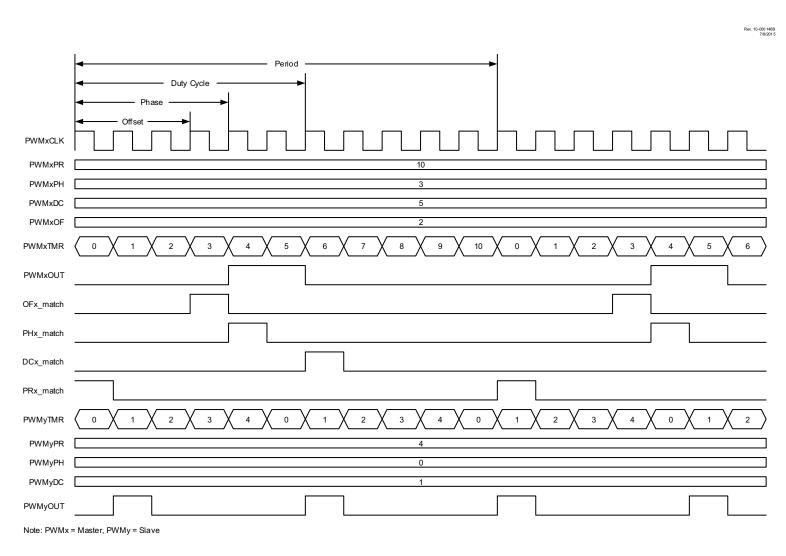
26.3.5 OFFSET MATCH IN CENTER ALIGNED MODE

When a master is operating in Center-Aligned mode the offset match event depends on which direction the PWMxTMR is counting. Clearing the OFO bit of the PWMxOFCON register will cause the OF_match event to occur when the timer is counting up. Setting the OFO bit of the PWMxOFCON register will cause the OF_match event to occur when the timer is counting down. The OFO bit is ignored in Non-Center-Aligned modes.

The OFO bit is double buffered and requires setting the LDA bit to take effect when the PWM module is operating.

Detailed timing diagrams of Center-Aligned mode using offset match control in Independent Slave with Sync Start mode can be seen in Figure 26-12 and Figure 26-13.

FIGURE 26-8: INDEPENDENT RUN MODE TIMING DIAGRAM



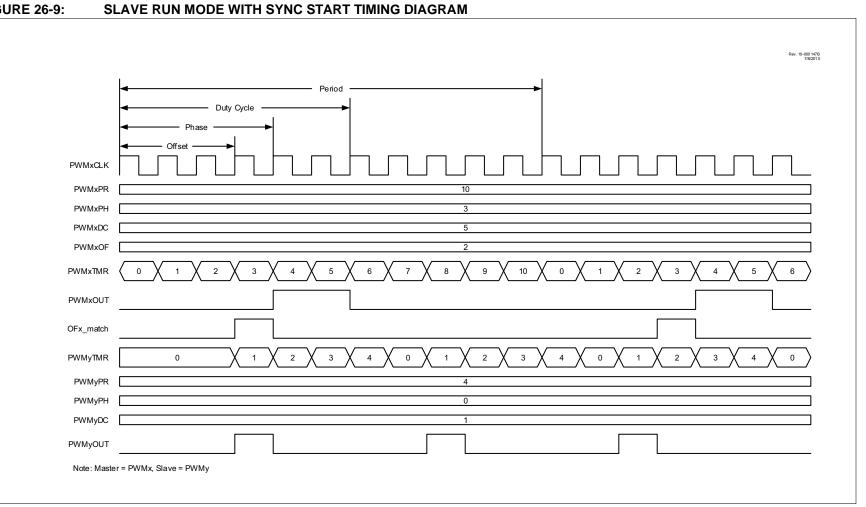
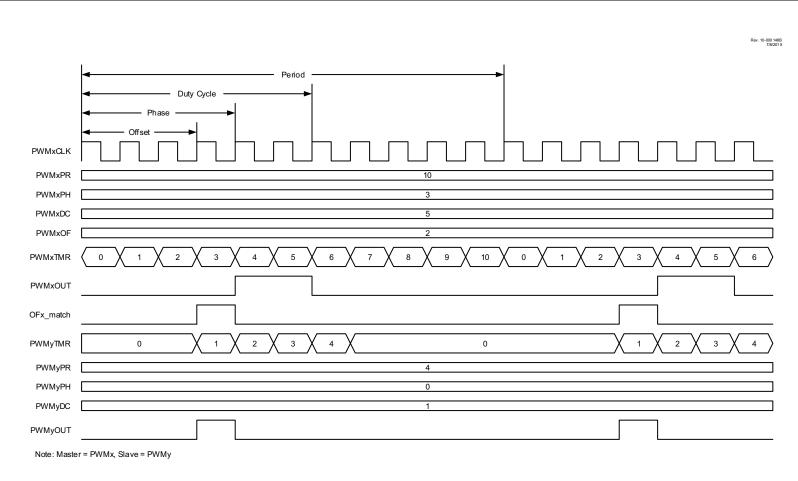


FIGURE 26-9:

FIGURE 26-10: ONE-SHOT SLAVE RUN MODE WITH SYNC START TIMING DIAGRAM



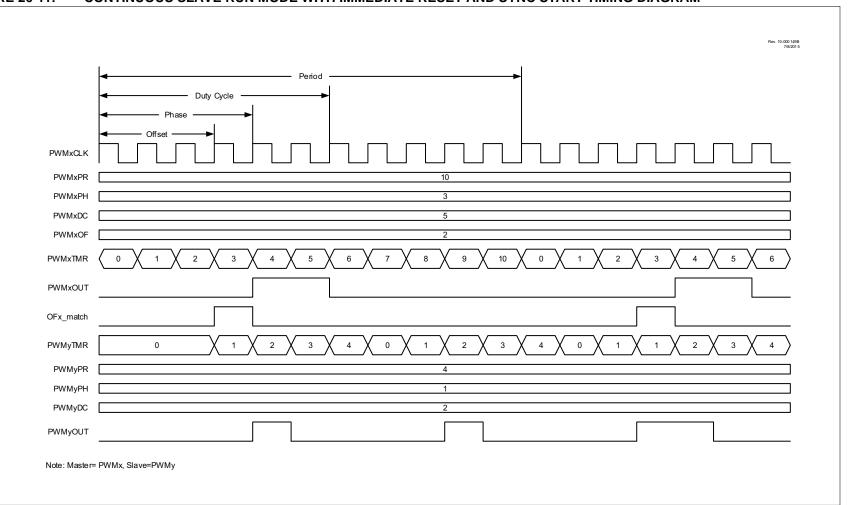
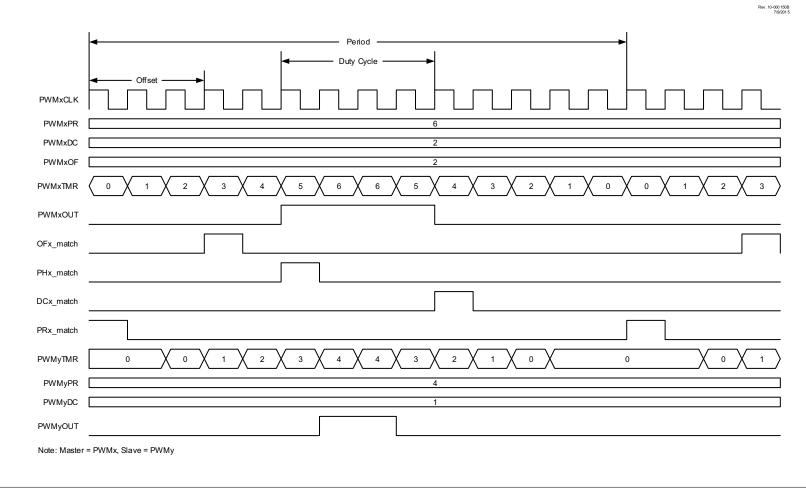


FIGURE 26-12: OFFSET MATCH ON INCREMENTING TIMER TIMING DIAGRAM



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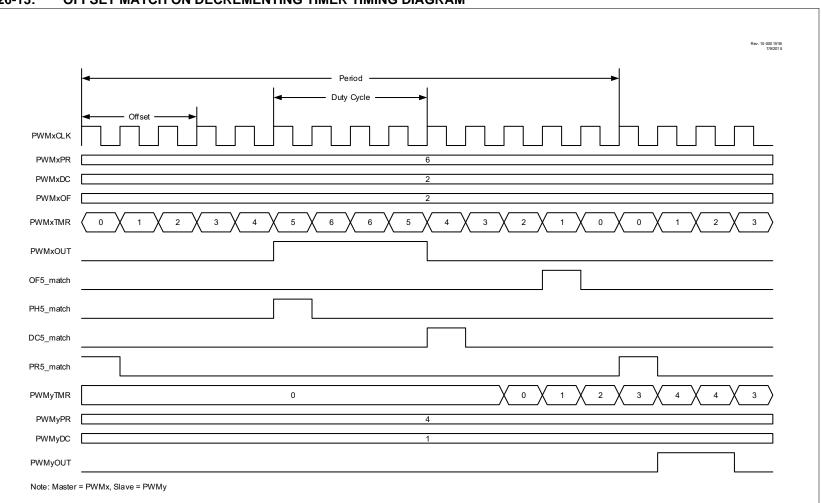


FIGURE 26-13: OFFSET MATCH ON DECREMENTING TIMER TIMING DIAGRAM

26.4 Reload Operation

Four of the PWM module control register pairs and one control bit are double buffered so that all can be updated simultaneously. These include:

- PWMxPHH:PWMxPHL register pair
- PWMxDCH:PWMxDCL register pair
- PWMxPRH:PWMxPRL register pair
- PWMxOFH:PWMxOFL register pair
- ODO control bit

When written to, these registers do not immediately affect the operation of the PWM. By default, writes to these registers will not be loaded into the PWM operating buffer registers until after the arming conditions are met. The arming control has two methods of operation:

- Immediate
- Triggered

The LDT bit of the PWMxLDCON register controls the arming method. Both methods require the LDA bit to be set. All four buffer pairs will load simultaneously at the loading event.

26.4.1 IMMEDIATE RELOAD

When the LDT bit is clear then the Immediate mode is selected and the buffers will be loaded at the first period event after the LDA bit is set. Immediate reloading is used when a PWM module is operating stand-alone or when the PWM module is operating as a master to other slave PWM modules.

26.4.2 TRIGGERED RELOAD

When the LDT bit is set then the Triggered mode is selected and a trigger event is required for the LDA bit to take effect. The trigger source is the buffer load event of one of the other PWM modules in the device. The triggering source is selected by the LDS<1:0> bits of the PWMxLDCON register. The buffers will be loaded at the first period event following the trigger event. Triggered reloading is used when a PWM module is operating as a slave to another PWM and it is necessary to synchronize the buffer reloads in both modules.

Note 1: The buffer load operation clears the LDA bit.

2: If the LDA bit is set at the same time as PWMxTMR = PWMxPR, the LDA bit is ignored until the next period event. Such is the case when triggered reload is selected and the triggering event occurs simultaneously with the target's period event

26.5 Operation in Sleep Mode

Each PWM module will continue to operate in Sleep mode when either the HFINTOSC or LFINTOSC is selected as the clock source by PWMxCLKCON<1:0>.

26.6 Interrupts

Each PWM module has four independent interrupts based on the phase, duty cycle, period, and offset match events. The interrupt flag is set on the rising edge of each of these signals. Refer to Figures 26-8 and 26-12 for detailed timing diagrams of the match signals.

26.7 Register Definitions: PWM Control

Long bit name prefixes for the 16-bit PWM peripherals are shown in Table 26-2. Refer to **Section 1.1 "Register and Bit naming conventions**" for more information

TABLE 26-2:

Peripheral	Bit Name Prefix
PWM5	PWM5
PWM6	PWM6
PWM11	PWM11
PWM12 ⁽¹⁾	PWM12

Note 1: PIC16(L)F1777/9 only.

REGISTER 26-1: PWMxCON: PWM CONTROL REGISTER

R/W-0/0	U-0	R/HS/HC-0/0	R/W-0/0	R/W-0/0	R/W-0/0	U-0	U-0
EN	—	OUT	POL	MODE<1:0>		_	—
bit 7							bit 0

Legend:HC = Bit is cleared by hardware<math>HS = Bit is set by hardwareR = Readable bitW = Writable bitU = Unimplemented bit, read as '0'<math>u = Bit is unchangedx = Bit is unknown-n/n = Value at POR and BOR/Value at all other Resets'1' = Bit is set'0' = Bit is cleared

bit 7	EN: PWM Module Enable bit 1 = Module is enabled 0 = Module is disabled
bit 6	Unimplemented: Read as '0'
bit 5	OUT: Output State of the PWM module
bit 4	 POL: PWM Output Polarity Control bit 1 = PWM output active state is low 0 = PWM output active state is high
bit 3-2	MODE<1:0>: PWM Mode Control bits 11 = Center Aligned mode 10 = Toggle On Match mode 01 = Set On Match mode 00 = Standard PWM mode
bit 1-0	Unimplemented: Read as '0'

U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
_	_	_	_	OFIE	PHIE	DCIE	PRIE
bit 7	·	•	•	·			bit 0
Legend:							
R = Readabl	le bit	W = Writable b	bit	U = Unimplem	ented bit, read a	as '0'	
u = Bit is und	changed	x = Bit is unkno	own	-n/n = Value at	POR and BOR	/Value at all oth	er Resets
'1' = Bit is se	et	'0' = Bit is clea	red				
bit 7-4	Unimplement	ed: Read as '0'					
bit 3	OFIE: Offset I	nterrupt Enable	bit				
		CPU on Offset Match					
		errupt CPU on (
bit 2		Interrupt Enable CPU on Phase N					
		terrupt CPU on I					
bit 1	DCIE: Duty C	vcle Interrupt Er	nable bit				
	1 = Interrupt (CPU on Duty Cy	cle Match				
	0 = Do not int	errupt CPU on I	Duty Cycle Mate	h			
bit 0 PRIE: Period Interrupt Enable bit							
		CPU on Period I errupt CPU on I					

REGISTER 26-2: PWMxINTE: PWM INTERRUPT ENABLE REGISTER

REGISTER 26-3: PWMxINTF: PWM INTERRUPT REQUEST REGISTER

U-0	U-0	U-0	U-0	R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0
_	_	—	—	OFIF	PHIF	DCIF	PRIF
bit 7							bit 0

Legend:		
HC = Bit is cleared by har	dware	HS = Bit is set by hardware
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-4	Unimplemented: Read as '0'
bit 3	OFIF : Offset Interrupt Flag bit ⁽¹⁾
	1 = Offset Match Event occurred
	0 = Offset Match Event did not occur
bit 2	PHIF: Phase Interrupt Flag bit ⁽¹⁾
	1 = Phase Match Event occurred
	0 = Phase Match Event did not occur
bit 1	DCIF: Duty Cycle Interrupt Flag bit ⁽¹⁾
	1 = Duty Cycle Match Event occurred
	0 = Duty Cycle Match Event did not occur
bit 0	PRIF: Period Interrupt Flag bit ⁽¹⁾
	1 = Period Match Event occurred
	0 = Period Match Event did not occur

Note 1: Bit is forced clear by hardware while module is disabled (EN = 0).

U-0	R/W-0/0	R/W-0/0	R/W-0/0	U-0	U-0	R/W-0/0	R/W-0/0	
_		PS<2:0>		_	_	CS<	1:0>	
bit 7					•		bit 0	
Legend:								
R = Readable	e bit	W = Writable bi	t	U = Unimpleme	ented bit, read as	s 'O'		
u = Bit is unc	hanged	x = Bit is unkno	wn	-n/n = Value at	POR and BOR/	/alue at all other	Resets	
'1' = Bit is set	t	'0' = Bit is clear	ed					
bit 7	Unimplemen	nted: Read as '0'						
bit 6-4	Unimplemented: Read as '0' PS<2:0>: Clock Source Prescaler Select bits 111 = Divide clock source by 128 110 = Divide clock source by 64 101 = Divide clock source by 32 100 = Divide clock source by 16 011 = Divide clock source by 8 010 = Divide clock source by 4 001 = Divide clock source by 2 000 = No Prescaler							
bit 3-2	Unimplemen	nted: Read as '0'						
bit 1-0	11 = Reserv 10 = LFINTC	CS<1:0>: Clock Source Select bits 11 = Reserved 10 = LFINTOSC (continues to operate during Sleep) 01 = HFINTOSC (continues to operate during Sleep)						

REGISTER 26-4: PWMxCLKCON: PWM CLOCK CONTROL REGISTER

PIC16(L)F1777/8/9

R/W/HC-0/0	R/W-0/0	U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0
LDA ⁽¹⁾	LDT	_	_	_	_	LDS<	1:0> (2)
bit 7	·						bit
Legend:							
R = Readable	bit	W = Writable b	it	U = Unimplem	ented bit, read a	is '0'	
u = Bit is unch	nanged	x = Bit is unkno	own	-n/n = Value at	POR and BOR	Value at all oth	er Resets
'1' = Bit is set		'0' = Bit is clea	red	HC = Cleared	by hardware		
bit 6	trigger oc: 0 = Do not loa If LDT = 0: $1 = Load the 00 = Do not loa LDT: Load Buf$	curs ad buffers, load ODO bit and OF ad buffers, load ffer on Trigger b	has completed x, PHx, DCx ar has completed t	and PRx buffers and PRx buffers a > bits to occur be	it the end of the	current period	ch the selecte
		-	d. Buffer loads a	are controlled by	/ the LDA bit alo	ne.	
bit 5-2	•	ed: Read as '0'	(2)				
bit 1-0	LDS<1:0>: Lo 10 = LD11_trig 01 = LD6_trig 00 = LD5_trig	ger	ce Select bits ⁽²⁾				
	nis bit is cleared b rent.	by the module af	ter a reload ope	eration. It can be	cleared in softw	are to clear an e	existing armin

REGISTER 26-5: PWMxLDCON: PWM RELOAD TRIGGER SOURCE SELECT REGISTER

2: The source corresponding to a PWM module's own LDx_trigger is reserved.

U-0	R/W-0/0	R/W-0/0	R/W-0/0	U-0	U-0	R/W-0/0	R/W-0/0
_	OF	M<1:0>	OFO ⁽¹⁾	_	_	OFS	<1:0>
bit 7							bit (
Legend:							
R = Readab	le bit	W = Writable t	pit	U = Unimplen	nented bit, read	as '0'	
u = Bit is un	changed	x = Bit is unkn	own	-n/n = Value a	t POR and BOF	R/Value at all oth	ner Resets
'1' = Bit is se	et	'0' = Bit is clea	ired				
bit 7	Unimpleme	nted: Read as '0'					
bit 6-5	OFM<1:0>:	Offset Mode Sele	ect bits				
	11 = Contin	uous Run Slave n	node with offset	triggered timer I	Reset and synch	nronized start	
	10 = One-sł	not Slave mode w	rith offset trigger	ed synchronized	Istart		
		Run mode with of	fset triggered sy	nchronized start	:		
		ndent Run mode					
bit 4		Match Output Co					
		0> = 11 (PWM ce					
		atch occurs when					
		atch occurs when			/n		
	this bit is ign	<u>0> = 00, 01, or 10</u> ored		<u>es)</u> .			
bit 3-2	0	nted: Read as '0'					
bit 1-0	•	Offset Trigger Sou					
	10 = OF11	00					
	01 = OF6_n	•					
	00 = OF5 n						

REGISTER 26-6: PWMxOFCON: PWM OFFSET TRIGGER SOURCE SELECT REGISTER

Note 1: The source corresponding to the PWM module's own OFx_match is reserved.

REGISTER 26-7:	PWMxPHH: PWMx PHASE COUNT HIGH REGISTER	
----------------	---	--

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
			PH<	:15:8>			
bit 7							bit 0
Legend:							
R = Readable	bit	W = Writable	bit	U = Unimpler	nented bit, read	d as '0'	
u = Bit is unch	anged	x = Bit is unkn	iown	-n/n = Value a	at POR and BC	R/Value at all	other Resets
'1' = Bit is set		'0' = Bit is clea	ared				

bit 7-0 **PH<15:8>**: PWM Phase High bits Upper eight bits of PWM phase count

REGISTER 26-8: PWMxPHL: PWMx PHASE COUNT LOW REGISTER

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
			PH<	7:0>			
bit 7							bit 0
-							
Legend:							

_ogona.		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **PH<7:0>**: PWM Phase Low bits Lower eight bits of PWM phase count

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
			DC<	15:8>			
bit 7							bit 0
Legend:							
R = Readable	bit	W = Writable	bit	U = Unimpler	nented bit, read	d as '0'	
u = Bit is unch	anged	x = Bit is unkn	nown	-n/n = Value a	at POR and BC	R/Value at all	other Resets
'1' = Bit is set		'0' = Bit is clea	ared				

REGISTER 26-9: PWMxDCH: PWMx DUTY CYCLE COUNT HIGH REGISTER

bit 7-0 DC<15:8>: PWM Duty Cycle High bits Upper eight bits of PWM duty cycle count

REGISTER 26-10: PWMxDCL: PWMx DUTY CYCLE COUNT LOW REGISTER

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
			DC<	7:0>			
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 DC<7:0>: PWM Duty Cycle Low bits Lower eight bits of PWM duty cycle count

REGISTER 26-11: PWMxPRH: PWMx PERIOD COUNT HIGH REGISTER

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
			PR<	15:8>			
bit 7							bit 0
Legend:							
R = Readable	bit	W = Writable I	oit	U = Unimpler	nented bit, read	d as '0'	
u = Bit is uncha	anged	x = Bit is unkn	own	-n/n = Value a	at POR and BC	R/Value at all	other Resets
'1' = Bit is set		'0' = Bit is clea	ared				

bit 7-0 **PR<15:8>**: PWM Period High bits Upper eight bits of PWM period count

REGISTER 26-12: PWMxPRL: PWMx PERIOD COUNT LOW REGISTER

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
			PR<	7:0>			
bit 7							bit 0
Legend:							
R = Readable	bit	W = Writable bi	it	U = Unimpler	nented bit, read	l as '0'	

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **PR<7:0>**: PWM Period Low bits Lower eight bits of PWM period count

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
			OF<	15:8>			
bit 7							bit 0
Legend:							
R = Readable	bit	W = Writable I	bit	U = Unimpler	nented bit, read	d as '0'	
u = Bit is unch	anged	x = Bit is unkn	iown	-n/n = Value a	at POR and BC	R/Value at all	other Resets
'1' = Bit is set		'0' = Bit is clea	ared				

REGISTER 26-13: PWMxOFH: PWMx OFFSET COUNT HIGH REGISTER

bit 7-0 **OF<15:8>**: PWM Offset High bits Upper eight bits of PWM offset count

REGISTER 26-14: PWMxOFL: PWMx OFFSET COUNT LOW REGISTER

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
OF<7:0>							
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **OF<7:0>**: PWM Offset Low bits Lower eight bits of PWM offset count

REGISTER 26-15: PWMxTMRH: PWMx TIMER HIGH REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0		
TMR<15:8>									
bit 7							bit 0		
Legend:									
R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'									
u = Bit is unchanged x = Bit is unknown			iown	-n/n = Value at POR and BOR/Value at all other Resets					
'1' = Bit is set		'0' = Bit is clea	ared						

bit 7-0 **TMR<15:8>**: PWM Timer High bits Upper eight bits of PWM timer counter

REGISTER 26-16: PWMxTMRL: PWMx TIMER LOW REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
			TMR<	<7:0>			
bit 7							bit 0
Legend:							

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **TMR<7:0>**: PWM Timer Low bits Lower eight bits of PWM timer counter Note: There are no long and short bit name variants for the following three mirror registers

REGISTER 26-17: PWMEN: PWMEN BIT MIRROR REGISTER

U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—		—	—	MPWM12EN ⁽¹⁾	MPWM11EN	MPWM6EN	MPWM5EN
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
'1' = Bit is set	'0' = Bit is cleared	-n/n = Value at POR and BOR/Value at all other Resets

bit 7-3	Unimplemented: Read as '0'

bit 2-0 **PWMxEN:** PWM11/PWM6/PWM5 Enable bits Mirror copy of each PWM module's PWMxCON<7> bit

Note 1: PIC16(L)F1777/9 only.

REGISTER 26-18: PWMLD: LDA BIT MIRROR REGISTER

U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	—	—	—	MPWM12LD ⁽¹⁾	MPWM11LD	MPWM6LD	MPWM5LD
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
'1' = Bit is set	'0' = Bit is cleared	-n/n = Value at POR and BOR/Value at all other Resets

bit 7-3	Unimplemented: Read as '0'
bit 2	MPWM11LD: PWM11LD bits Mirror copy of each PWM module's PWMxLDCON<7> bit
bit 1	MPWM6LD: PWM6LD bits Mirror copy of each PWM module's PWMxLDCON<7> bit
bit 0	MPWM5LD: PWM5LD bits Mirror copy of each PWM module's PWMxLDCON<7> bit

Note 1: PIC16(L)F1777/9 only.

REGISTER 26-19: PWMOUT: PWMOUT BIT MIRROR REGISTER

U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
_	—	_	_	MPWM12OUT ⁽¹⁾	MPWM11OUT	MPWM6OUT	MPWM5OUT
bit 7							bit 0

Legend:		
R = Readable bi	t W = Writable bit	U = Unimplemented bit, read as '0'
'1' = Bit is set	'0' = Bit is cleared	-n/n = Value at POR and BOR/Value at all other Resets
bit 7-3	Unimplemented: Read as '0'	
bit 2	MPWM11OUT: PWM11 OUT bits Mirror copy of each PWM module's PWMxCON	<5> bit
bit 1	MPWM6OUT: PWM6 OUT bits Mirror copy of each PWM module's PWMxCON	<5> bit
bit 0	MPWM5OUT: PWM5 OUT bits Mirror copy of each PWM module's PWMxCON	<5> bit

Note 1: PIC16(L)F1777/9 only.

PIC16(L)F1777/8/9

TABLE 26-3: SUMMARY OF REGISTERS ASSOCIATED WITH PWM

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page		
OSCCON	SPLLEN	N IRCF<3:0>			—	SCS	<1:0>	116			
PWMEN	_	_	_	_	MPWM12EN ⁽¹⁾	MPWM11EN	MPWM6EN	MPWM5EN	355		
PWMLD	_	—	—	_	MPWM12LD ⁽¹⁾	MPWM11LD	MPWM6LD	MPWM5LD	355		
PWMOUT	—	_	_	_	MPWM12OUT ⁽¹⁾	MPWM11OUT	MPWM6OUT	MPWM5OUT	355		
PWM5PHL	PH<7:0>										
PWM5PHH	PH<15:8>										
PWM5DCL	DC<7:0>										
PWM5DCH	DC<15:8>										
PWM5PRL	PR<7:0>										
PWM5PRH	PR<15:8>										
PWM5OFL	OF<7:0>										
PWM50FH					OF<15:8>				353		
PWM5TMRL					TMR<7:0>				354		
PWM5TMRH					TMR<15:8>				354		
PWM5CON	EN	_	OUT	POL	MODE	<1:0>	_	_	345		
PWM5INTE		_	_	_	OFIE	PHIE	DCIE	PRIE	346		
PWM5INTF	_	_	_	_	OFIE	PHIF	DCIF	PRIF	346		
PWM5CLKCON			PS<2:0>		_	_		:1:0>	347		
PWM5LDCON	LDA	LDT		_	_			<1:0>	348		
PWM50FC0N			<1:0>	OEO					349		
PWM6PHL	- OFM<1:0> OFO - OFS<1:0>										
PWM6PHL PWM6PHH	PH<7:0> PH<15:8>										
PWM6DCL											
PWM6DCL PWM6DCH	DC<7:0>										
PWM6PRL	DC<15:8>										
PWM6PRL PWM6PRH	PR<7:0>										
PWM60FL	PR<15:8>										
	OF<7:0>										
PWM60FH	OF<15:8>										
PWM6TMRL					TMR<7:0>				354		
PWM6TMRH			OUT	DOI	TMR<15:8>	-1.0			354		
PWM6CON	EN	_	OUT	POL		MODE<1:0>			345		
PWM6INTE	_	_		_	OFIE	PHIE	DCIE	PRIE	346		
PWM6INTF	_	_	—	—	OFIF	PHIF	DCIF	PRIF	346		
PWM6CLKCON	-	1.57	PS<2:0>		-			<1:0>	347		
PWM6LDCON	LDA	LDT	—	-	_			<1:0>	348		
PWM60FCON	- OFM<1:0> OFO - OFS<1:0>								349		
PWM11PHL					PH<7:0>				350		
PWM11PHH	PH<15:8>								350		
PWM11DCL					DC<7:0>				351		
PWM11DCH					DC<15:8>				351		
PWM11PRL					PR<7:0>				352		
PWM11PRH					PR<15:8>				352		
PWM110FL					OF<7:0>				353		
PWM110FH					OF<15:8>				353		
PWM11TMRL	TMR<7:0>								354		
PWM11TMRH	TMR<15:8>								354		
PWM11CON	EN	—	OUT	POL	MODE	<1:0>	—	—	345		
PWM11INTE	—	—	_	_	OFIE	PHIE	DCIE	PRIE	346		
PWM11INTF	—	—	—	—	OFIF	PHIF	DCIF	PRIF	346		
PWM11CLKCON	- PS<2:0> CS<1:0>						347				
PWM11LDCON	LDA LDT LDS<1:0>										

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by PWM. Note 1: PIC16(L)F1777/9 only.

TADLE 20-3.												
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page			
PWM110FCON	- OFM<1:0> OFO - OFS<1:0>											
PWM12PHL ⁽¹⁾	PH<7:0>											
PWM12PHH ⁽¹⁾	PH<15:8>											
PWM12DCL ⁽¹⁾		DC<7:0>										
PWM12DCH ⁽¹⁾		DC<15:8>										
PWM12PRL ⁽¹⁾		PR<7:0>										
PWM12PRH ⁽¹⁾		PR<15:8>										
PWM12OFL ⁽¹⁾	OF<7:0>											
PWM12OFH ⁽¹⁾	OF<15:8>											
PWM12TMRL ⁽¹⁾	TMR<7:0>											
PWM12TMRH ⁽¹⁾	TMR<15:8>											
PWM12CON ⁽¹⁾	EN	_	OUT	POL	MODE<1:0> — —				345			
PWM12INTE ⁽¹⁾	_	_	_	_	OFIE	PHIE	DCIE	PRIE	346			
PWM12INTF ⁽¹⁾	_	_	_	_	OFIF	PHIF	DCIF	PRIF	346			
PWM12CLKCON ⁽¹⁾	- PS<2:0> CS<1:0>											
PWM12LDCON ⁽¹⁾	LDA	LDT – – – LDS<1:0>					<1:0>	348				
PWM12OFCON ⁽¹⁾	_	OFM<1:0> OFO — — OFS<1:0>										

TABLE 26-3: SUMMARY OF REGISTERS ASSOCIATED WITH PWM (CONTINUED)

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by PWM.

Note 1: PIC16(L)F1777/9 only.

TABLE 26-4: SUMMARY OF CONFIGURATION WORDS WITH CLOCK SOURCES

Name	Bits	Bit -/7	Bit -/6	Bit 13/5	Bit 12/4	Bit 11/3	Bit 10/2	Bit 9/1	Bit 8/0	Register on Page
0015104	13:8	_	—	FCMEN	IESO	CLKOUTEN	BOREN<1:0>		—	05
CONFIG1	7:0	CP	MCLRE	PWRTE	WDTE<1:0>		FOSC<2:0>			95

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by clock sources.

PIC16(L)F1777/8/9

NOTES:

27.0 COMPLEMENTARY OUTPUT GENERATOR (COG) MODULES

The primary purpose of the Complementary Output Generator (COG) is to convert a single-output PWM signal into a two-output complementary PWM signal. The COG can also convert two separate input events into a single or complementary PWM output.

Device	COG1	COG2	COG3	COG4
PIC16(L)F1778	•	•	•	
PIC16(L)F1777/9	•	•	•	•

The COG PWM frequency and duty cycle are determined by a rising event input and a falling event input. The rising event and falling event may be the same source. Sources may be synchronous or asynchronous to the COG clock.

The rate at which the rising event occurs determines the PWM frequency. The time from the rising event to the falling event determines the duty cycle.

A selectable clock input is used to generate the phase delay, blanking, and dead-band times. Dead-band time can also be generated with a programmable delay chain, which is independent from all clock sources.

Simplified block diagrams of the various COG modes are shown in Figure 27-2 through Figure 27-6.

The COG module has the following features:

- Six modes of operation:
 - Steered PWM mode
 - Synchronous Steered PWM mode
 - Forward Full-Bridge mode
 - Reverse Full-Bridge mode
 - Half-Bridge mode
 - Push-Pull mode
- Selectable COG_clock clock source
- Independently selectable rising event sources
- Independently selectable falling event sources
- Independently selectable edge or level event sensitivity
- Independent output polarity selection
- Phase delay with independent rising and falling delay times
- · Dead-band control with:
 - independent rising and falling event dead-band times
 - Synchronous and asynchronous timing
- Blanking control with independent rising and falling event blanking times
- Auto-shutdown control with:
 - Independently selectable shutdown sources
 - Auto-restart enable
 - Auto-shutdown pin override control (high, low, off, and High-Z)

27.1 Output to Pins (all modes)

The COG peripheral has four outputs: COGA, COGB, COGC, and COGD.

The operating mode, selected with the MD<2:0> bits of the COGxCON0 register, determine the waveform available at each output. An individual peripheral source control for each device pin selects the pin or pins at which the outputs will appear. Please refer to the RxyPPS register (Register 12-2) for more information.

27.2 Event-Driven PWM (All Modes)

Besides generating PWM and complementary outputs from a single PWM input, the COG can also generate PWM waveforms from a periodic rising event and a separate falling event. In this case, the falling event is usually derived from analog feedback within the external PWM driver circuit. In this configuration, high-power switching transients may trigger a false falling event that needs to be blanked out. The COG can be configured to blank falling (and rising) event inputs for a period of time immediately following the rising (and falling) event drive output. This is referred to as input blanking and is covered in Section 27.8 "Blanking Control".

It may be necessary to guard against the possibility of external circuit faults. In this case, the active drive must be terminated before the Fault condition causes damage. This is referred to as auto-shutdown and is covered in Section 27.10 "Auto-Shutdown Control".

The COG can be configured to operate in phase delayed conjunction with another PWM. The active drive cycle is delayed from the rising event by a phase delay timer. Phase delay is covered in more detail in **Section 27.9 "Phase Delay"**.

A typical operating waveform, with phase delay and dead band, generated from a single CCP1 input is shown in Figure 27-10.

27.3 Modes of Operation

27.3.1 STEERED PWM MODES

In Steered PWM mode, the PWM signal derived from the input event sources is output as a single phase PWM which can be steered to any combination of the four COG outputs. Output steering takes effect on the instruction cycle following the write to the COGxSTR register.

Synchronous Steered PWM mode is identical to the Steered PWM mode except that changes to the output steering take effect on the first rising event after the COGxSTR register write. Static output data is not synchronized.

Steering mode configurations are shown in Figure 27-2 and Figure 27-3.

Steered PWM and Synchronous Steered PWM modes are selected by setting the MD<2:0> bits of the COGxCON0 register (Register 27-1) to '000' and '001', respectively.

27.3.2 FULL-BRIDGE MODES

In both Forward and Reverse Full-Bridge modes, two of the four COG outputs are active and the other two are inactive. Of the two active outputs, one is modulated by the PWM input signal and the other is on at 100% duty cycle. When the direction is changed, the dead-band time is inserted to delay the modulated output. This gives the unmodulated driver time to shut down, thereby, preventing shoot-through current in the series connected power devices.

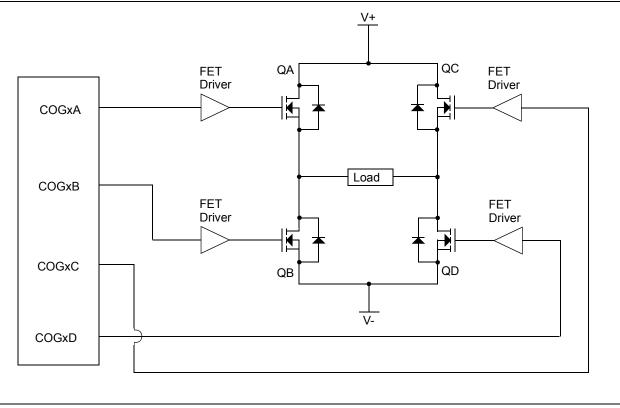
In Forward Full-Bridge mode, the PWM input modulates the COGxD output and drives the COGA output at 100%.

In Reverse Full-Bridge mode, the PWM input modulates the COGxB output and drives the COGxC output at 100%.

The full-bridge configuration is shown in Figure 27-4. Typical full-bridge waveforms are shown in Figure 27-12 and Figure 27-13.

Full-Bridge Forward and Full-Bridge Reverse modes are selected by setting the MD<2:0> bits of the COGxCON0 register to '010' and '011', respectively.





27.3.3 HALF-BRIDGE MODE

In Half-Bridge mode, the COG generates a two output complementary PWM waveform from rising and falling event sources. In the simplest configuration, the rising and falling event sources are the same signal, which is a PWM signal with the desired period and duty cycle. The COG converts this single PWM input into a dual complementary PWM output. The frequency and duty cycle of the dual PWM output match those of the single input PWM signal. The off-to-on transition of each output can be delayed from the on-to-off transition of the other output, thereby, creating a time immediately after the PWM transition where neither output is driven. This is referred to as dead-band time and is covered in **Section 27.7 "Dead-Band Control"**.

The half-bridge configuration is shown in Figure 27-5. A typical operating waveform, with dead band, generated from a single CCP1 input is shown in Figure 27-9.

The primary output is available on either, or both, COGxA and COGxC. The complementary output is available on either, or both, COGxB and COGxD.

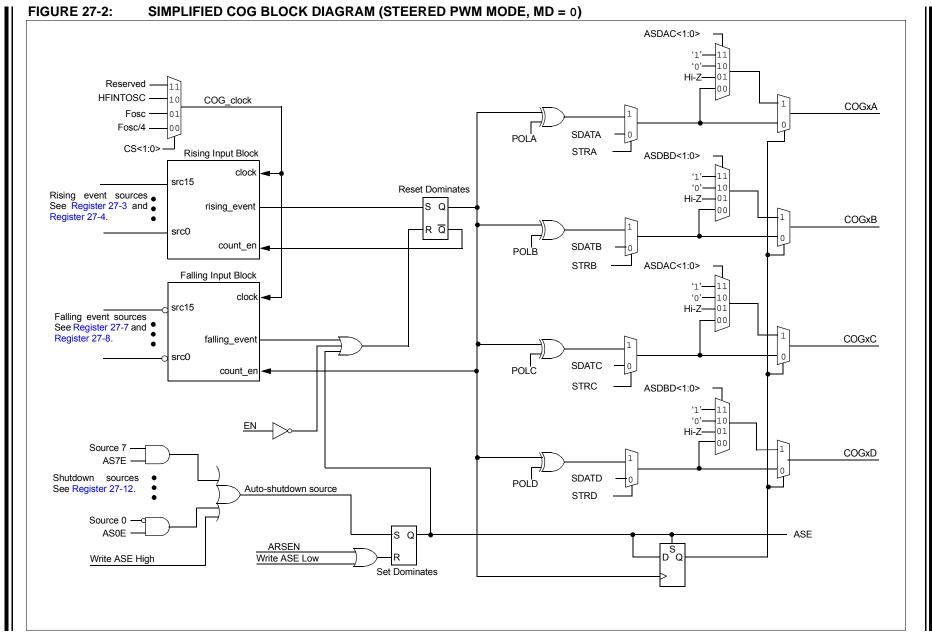
Half-Bridge mode is selected by setting the MD<2:0> bits of the COGxCON0 register to '100'.

27.3.4 PUSH-PULL MODE

In Push-Pull mode, the COG generates a single PWM output that alternates between the two pairs of the COG outputs at every PWM period. COGxA has the same signal as COGxC. COGxB has the same signal as COGxD. The output drive activates with the rising input event and terminates with the falling event input. Each rising event starts a new period and causes the output to switch to the COG pair not used in the previous period.

The Push-Pull configuration is shown in Figure 27-6. A typical Push-Pull waveform generated from a single CCP1 input is shown in Figure 27-11.

Push-Pull mode is selected by setting the MD<2:0> bits of the COGxCON0 register to '101'.



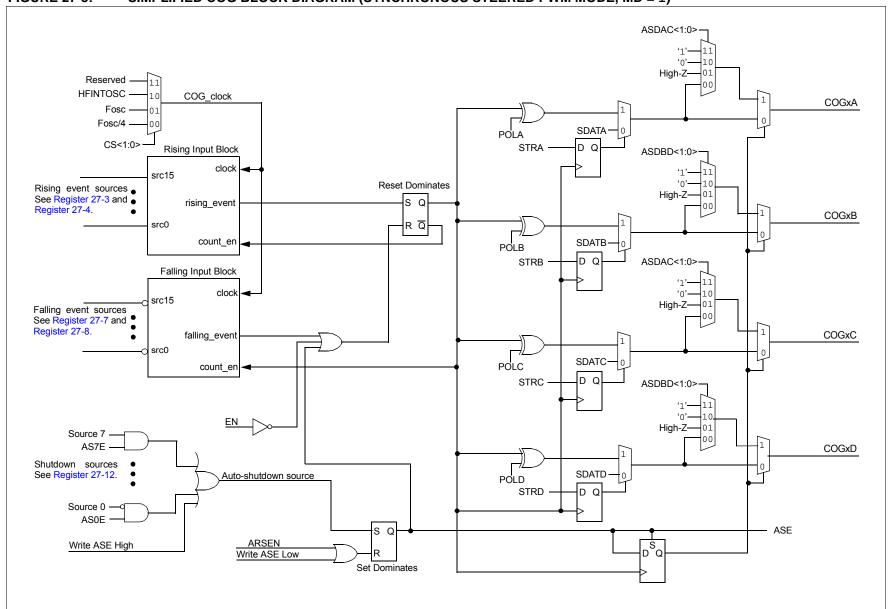
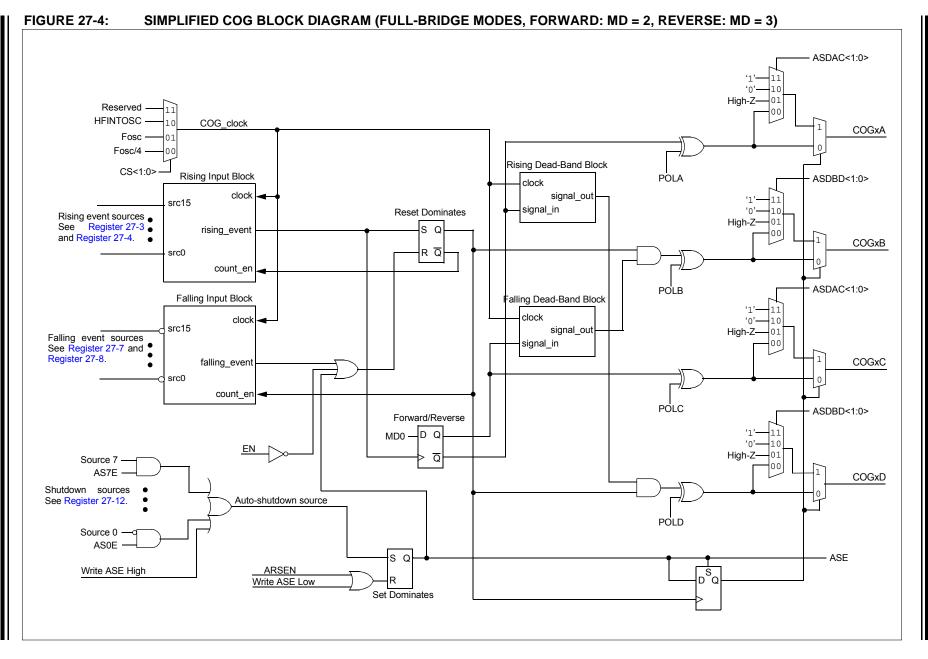


FIGURE 27-3: SIMPLIFIED COG BLOCK DIAGRAM (SYNCHRONOUS STEERED PWM MODE, MD = 1)



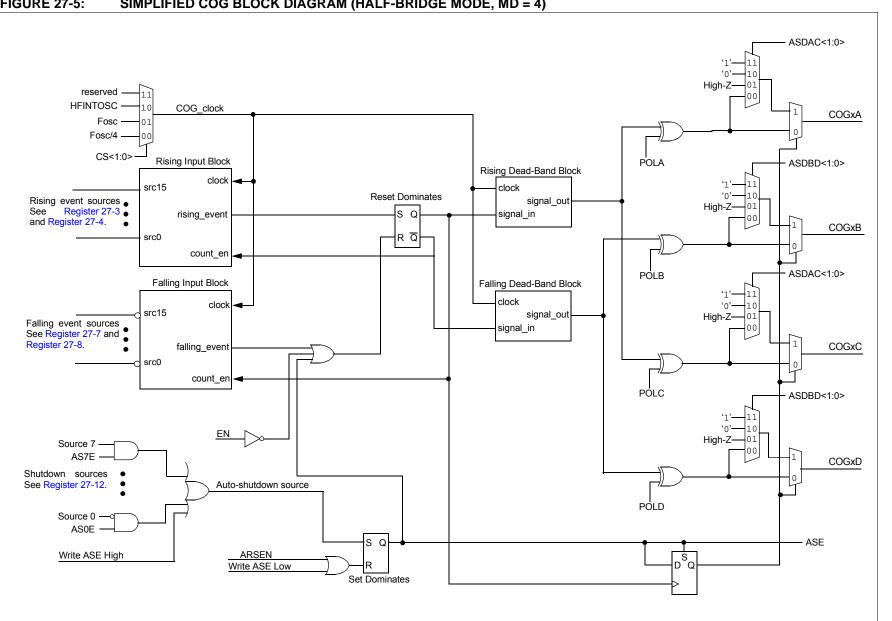
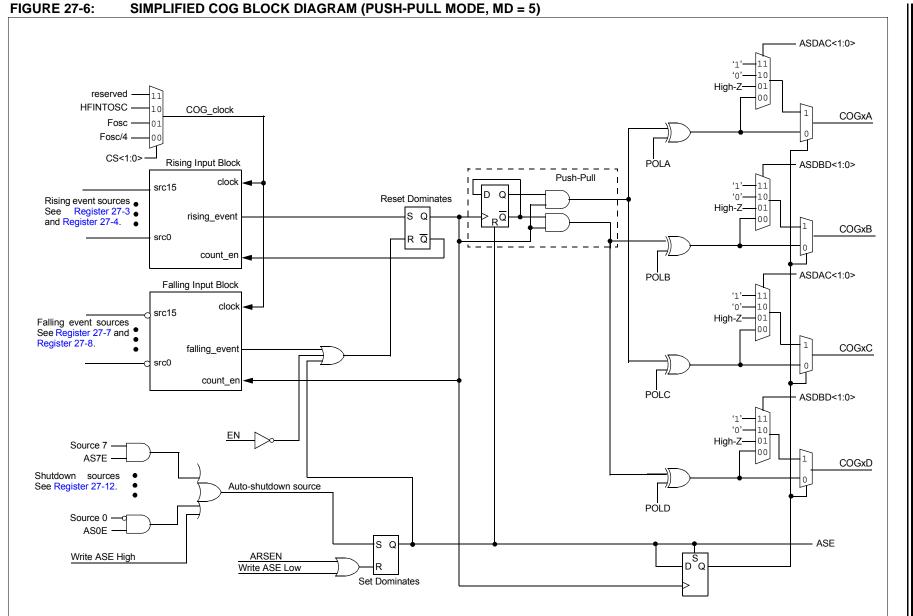
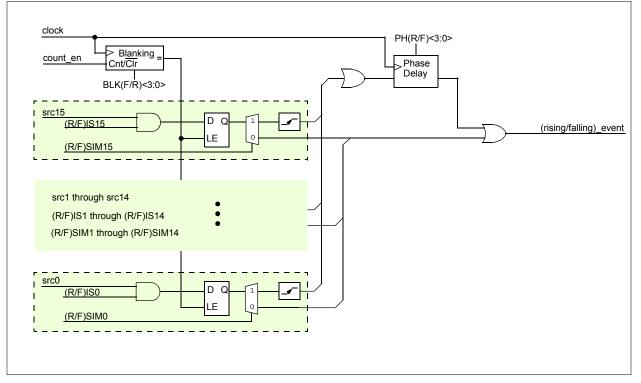


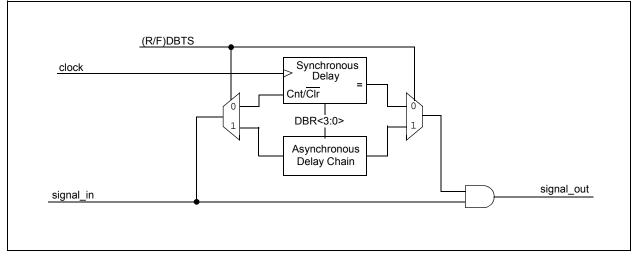
FIGURE 27-5: SIMPLIFIED COG BLOCK DIAGRAM (HALF-BRIDGE MODE, MD = 4)











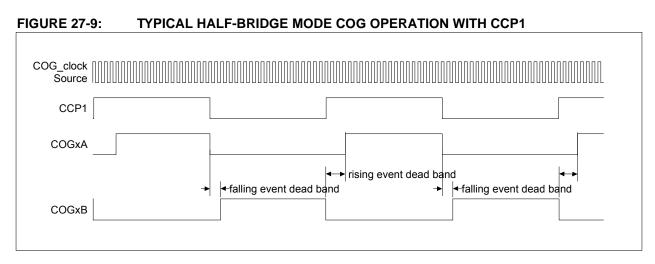


FIGURE 27-10: HALF-BRIDGE MODE COG OPERATION WITH CCP1 AND PHASE DELAY

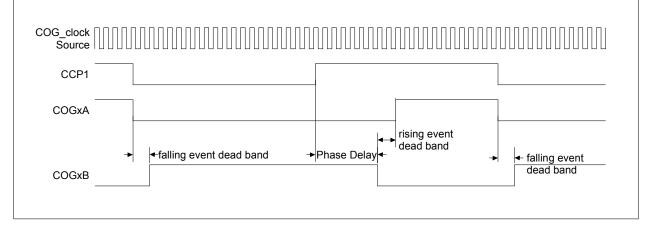


FIGURE 27-11: PUSH-PULL MODE COG OPERATION WITH CCP1

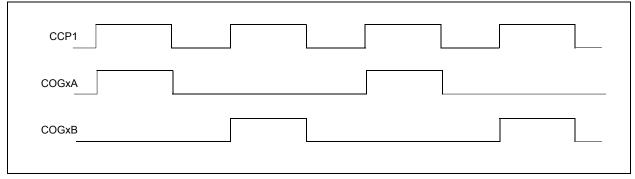


FIGURE 27-12:	FULL-BRIDGE FORWARD MODE COG OPERATION WITH CCP1					
CCP1						
COGxA						
COGxB						
COGxC						
COGxD						

FIGURE 27-13: FULL-BRIDGE MODE COG OPERATION WITH CCP1 AND DIRECTION CHANGE

CCP1			
COGxA			
COGxB	 → ← f 	alling event dead band	
COGxC			
COGxD	I		
MD0			

27.4 Clock Sources

The COG_clock is used as the reference clock to the various timers in the peripheral. Timers that use the COG_clock include:

- Rising and falling dead-band time
- Rising and falling blanking time
- Rising and falling event phase delay

Clock sources available for selection include:

- 16 MHz HFINTOSC (active during Sleep)
- Instruction clock (Fosc/4)
- System clock (Fosc)

The clock source is selected with the CS<1:0> bits of the COGxCON0 register (Register 27-1).

27.5 Selectable Event Sources

The COG uses any combination of independently selectable event sources to generate the complementary waveform. Sources fall into two categories:

- · Rising event sources
- · Falling event sources

The rising event sources are selected by setting bits in the COGxRIS0 and COGxRIS1 registers (Register 27-3 and Register 27-4). The falling event sources are selected by setting bits in the COGxFIS0 and COGxF1 registers (Register 27-7 and Register 27-8). All selected sources are OR'd together to generate the corresponding event signal. Refer to Figure 27-7.

27.5.1 EDGE VS. LEVEL SENSING

Event input detection may be selected as level or edge sensitive. The detection mode is individually selectable for every source. Rising source detection modes are selected with the COGxRSIM0 and COGxRSIM1 registers (Register 27-5 and Register 27-6). Falling source detection modes are selected with the COGxFSIM0 and COGxFSIM1 registers (Register 27-9 and Register 27-10). A set bit selects edge detection for the corresponding event source. A cleared bit selects level detection.

In general, events that are driven from a periodic source should be edge detected and events that are derived from voltage thresholds at the target circuit should be level sensitive. Consider the following two examples:

1. The first example is an application in which the period is determined by a 50% duty cycle clock on the rising event input and the COG output duty cycle is determined by a voltage level fed back through a comparator on the falling event input. If the clock input is level sensitive, duty cycles less than 50% will exhibit erratic operation because the level sensitive clock will suppress the comparator feedback.

2. The second example is similar to the first except that the duty cycle is close to 100%. The feedback comparator high-to-low transition trips the COG drive off, but almost immediately the period source turns the drive back on. If the off cycle is short enough, the comparator input may not reach the low side of the hysteresis band precluding an output change. The comparator output stays low and without a high-to-low transition to trigger the edge sense, the drive of the COG output will be stuck in a constant drive-on condition. See Figure 27-14.

Rising (CCP1)
Falling (C1OUT)
C1IN- hyst
COGOUT
Edge Sensitive
Rising (CCP1)
Falling (C1OUT)
C1IN- hyst
COGOUT
Level Sensitive

FIGURE 27-14: EDGE VS. LEVEL SENSE

27.5.2 RISING EVENT

The rising event starts the PWM output active duty cycle period. The rising event is the low-to-high transition of the rising_event output. When the rising event phase delay and dead-band time values are zero, the primary output starts immediately. Otherwise, the primary output is delayed. The rising event source causes all the following actions:

- Start rising event phase delay counter (if enabled).
- Clear complementary output after phase delay.
- Start falling event input blanking (if enabled).
- Start dead-band delay (if enabled).
- · Set primary output after dead-band delay expires.

27.5.3 FALLING EVENT

The falling event terminates the PWM output active duty cycle period. The falling event is the high-to-low transition of the falling_event output. When the falling event phase delay and dead-band time values are zero, the complementary output starts immediately. Otherwise, the complementary output is delayed. The falling event source causes all the following actions:

- Start falling event phase delay counter (if enabled).
- · Clear primary output.
- Start rising event input blanking (if enabled).
- · Start falling event dead-band delay (if enabled).
- Set complementary output after dead-band delay expires.

27.6 Output Control

Upon disabling, or immediately after enabling the COG module, the primary COG outputs are inactive and complementary COG outputs are active.

27.6.1 OUTPUT ENABLES

There are no output enable controls in the COG module. Instead, each device pin has an individual output selection control called the PPS register. All four COG outputs are available for selection in the PPS register of every pin.

When a COG output is enabled by PPS selection, the output on the pin has several possibilities which depend on the mode, steering control, EN bit, and shutdown state as shown in Table 27-2 and Table 27-3.

TABLE 27-2:PIN OUTPUT STATESMD<2:0> = 00x

EN	STR bit	Shutdown	Output
x	0	Inactive	Static steering data
x	1	Active	Shutdown override
0	1	Inactive	Inactive state
1	1	Inactive	Active PWM signal

TABLE 27-3: PIN OUTPUT STATES MD<2:0>> 001

EN	STR bit	Shutdown	Output
х	x	Inactive	Inactive state
x	x	Active	Shutdown override
1	x	Inactive	Active PWM signal

27.6.2 POLARITY CONTROL

The polarity of each COG output can be selected independently. When the output polarity bit is set, the corresponding output is active-low. Clearing the output polarity bit configures the corresponding output as active-high. However, polarity affects the outputs in only one of the four shutdown override modes. See **Section 27.10** "Auto-Shutdown Control" for more details.

Output polarity is selected with the POLA through POLD bits of the COGxCON1 register (Register 27-2).

27.7 Dead-Band Control

The dead-band control provides for non-overlapping PWM output signals to prevent shoot-through current in the external power switches. Dead-band time affects the output only in the Half-Bridge mode and when changing direction in the Full-Bridge mode.

The COG contains two dead-band timers. One dead-band timer is used for rising event dead-band control. The other is used for falling event dead-band control. Timer modes are selectable as either:

- · Asynchronous delay chain
- Synchronous counter

The Dead-Band Timer mode is selected for the rising event and falling event dead-band times with the respective RDBS and FDBS bits of the COGxCON1 register (Register 27-2).

In Half-Bridge mode, the rising event dead-band time delays all selected primary outputs from going active for the selected dead-band time after the rising event. COGxA and COGxC are the primary outputs in Half-Bridge mode.

In Half-Bridge mode, the falling event dead-band time delays all selected complementary outputs from going active for the selected dead-band time after the falling event. COGxB and COGxD are the complementary outputs in Half-Bridge mode.

In Full-Bridge mode, the dead-band delay occurs only during direction changes. The modulated output is delayed for the falling event dead-band time after a direction change from forward to reverse. The modulated output is delayed for the rising event dead-band time after a direction change from reverse to forward.

27.7.1 ASYNCHRONOUS DELAY CHAIN DEAD-BAND DELAY

Asynchronous dead-band delay is determined by the time it takes the input to propagate through a series of delay elements. Each delay element is a nominal five nanoseconds.

For rising event asynchronous dead-band delay set the RDBS bit of the COGxCON0 register and set the COGxDBR register (Register 27-14) value to the desired number of delay elements in the rising event dead-band time.

For falling event asynchronous dead-band delay set the FDBS bit of the COGxCON0 register and set the COGxDBF register (Register 27-15) value to the desired number of delay elements in the falling event dead-band time.

Setting the value to zero disables dead-band delay.

27.7.2 SYNCHRONOUS COUNTER DEAD-BAND DELAY

Synchronous counter dead band is timed by counting COG_clock periods from zero up to the value in the dead-band count register. Use Equation 27-1 to calculate dead-band times.

For rising event synchronous dead-band delay clear the RDBS bit of the COGxCON0 register and set the COGxDBR count register value to the number of COG_clock periods in the rising event dead-band time.

For falling event synchronous dead-band delay clear the FDBS bit of the COGxCON0 register and set the COGxDBF count register value to the number of COG_clock periods in the falling event dead-band time.

When the value is zero, dead-band delay is disabled.

27.7.3 SYNCHRONOUS COUNTER DEAD-BAND TIME UNCERTAINTY

When the rising and falling events that trigger the dead-band counters come from asynchronous inputs, it creates uncertainty in the synchronous counter dead-band time. The maximum uncertainty is equal to one COG_clock period. Refer to Example 27-1 for more detail.

When event input sources are asynchronous with no phase delay, use the Asynchronous Delay Chain Dead-Band mode to avoid the dead-band time uncertainty.

27.7.4 RISING EVENT DEAD BAND

Rising event dead band delays the turn-on of the primary outputs from when complementary outputs are turned off. The rising event dead-band time starts when the rising_ event output goes true.

See Section 27.7.1 "Asynchronous Delay Chain Dead-Band Delay" and Section 27.7.2 "Synchronous Counter Dead-Band Delay" for more information on setting the rising edge dead-band time.

27.7.5 FALLING EVENT DEAD BAND

Falling event dead band delays the turn-on of complementary outputs from when the primary outputs are turned off. The falling event dead-band time starts when the falling_event output goes true.

See Section 27.7.1 "Asynchronous Delay Chain Dead-Band Delay" and Section 27.7.2 "Synchronous Counter Dead-Band Delay" for more information on setting the rising edge dead-band time.

27.7.6 DEAD-BAND OVERLAP

There are two cases of potential dead-band overlap:

- · Rising-to-falling
- · Falling-to-rising

27.7.6.1 Rising-to-Falling Overlap

In this case, the falling event occurs while the rising event dead-band counter is still counting. When this happens, the primary drives are suppressed and the dead band extends by the falling event dead-band time. At the termination of the extended dead-band time, the complementary drive goes true.

27.7.6.2 Falling-to-Rising Overlap

In this case, the rising event occurs while the falling event dead-band counter is still counting. When this happens, the complementary drive is suppressed and the dead band extends by the rising event dead-band time. At the termination of the extended dead-band time, the primary drive goes true.

27.8 Blanking Control

Input blanking is a function whereby the event inputs can be masked or blanked for a short period of time. This is to prevent electrical transients caused by the turn-on/off of power components from generating a false input event.

The COG contains two blanking counters: one triggered by the rising event and the other triggered by the falling_event. The counters are cross coupled with the events they are blanking. The falling event blanking counter is used to blank rising input events and the rising event blanking counter is used to blank falling input events. Once started, blanking extends for the time specified by the corresponding blanking counter.

Blanking is timed by counting COG_clock periods from zero up to the value in the blanking count register. Use Equation 27-1 to calculate blanking times.

27.8.1 FALLING EVENT BLANKING OF RISING EVENT INPUTS

The falling event blanking counter inhibits rising event inputs from triggering a rising event. The falling event blanking time starts when the rising_event output drive goes false.

The falling event blanking time is set by the value contained in the COGxBLKF register (Register 27-17). Blanking times are calculated using the formula shown in Equation 27-1.

When the COGxBLKF value is zero, falling event blanking is disabled and the blanking counter output is true, thereby, allowing the event signal to pass straight through to the event trigger circuit.

27.8.2 RISING EVENT BLANKING OF FALLING EVENT INPUTS

The rising event blanking counter inhibits falling event inputs from triggering a falling event. The rising event blanking time starts when the falling_event output drive goes false.

The rising event blanking time is set by the value contained in the COGxBLKR register (Register 27-16).

When the COGxBLKR value is zero, rising event blanking is disabled and the blanking counter output is true, thereby, allowing the event signal to pass straight through to the event trigger circuit.

27.8.3 BLANKING TIME UNCERTAINTY

When the rising and falling sources that trigger the blanking counters are asynchronous to the COG_clock, it creates uncertainty in the blanking time. The maximum uncertainty is equal to one COG_clock period. Refer to Equation 27-1 and Example 27-1 for more detail.

27.9 Phase Delay

It is possible to delay the assertion of either or both the rising event and falling events. This is accomplished by placing a non-zero value in COGxPHR or COGxPHF phase-delay count registers, respectively (Register 27-18 and Register 27-19). Refer to Figure 27-10 for COG operation with CCP1 and phase delay. The delay from the input rising event signal switching to the actual assertion of the events is calculated the same as the dead-band and blanking delays. Refer to Equation 27-1.

When the phase-delay count value is zero, phase delay is disabled and the phase-delay counter output is true, thereby, allowing the event signal to pass straight through to the complementary output driver flop.

27.9.1 CUMULATIVE UNCERTAINTY

It is not possible to create more than one COG_clock of uncertainty by successive stages. Consider that the phase-delay stage comes after the blanking stage, the dead-band stage comes after either the blanking or phase-delay stages, and the blanking stage comes after the dead-band stage. When the preceding stage is enabled, the output of that stage is necessarily synchronous with the COG_clock, which removes any possibility of uncertainty in the succeeding stage.

EQUATION 27-1: PHASE, DEAD-BAND, AND BLANKING TIME CALCULATION

$T_{\min} = \frac{\text{Count}}{F_{COG_\text{clow}}}$	 ck				
$T_{\max} = \frac{\text{Count} + 1}{F_{COG_clock}}$					
$T_{\text{uncertainty}} = T_{\text{ma}}$	$x - T_{min}$				
Also:					
$T_{\text{uncertainty}} = \frac{1}{F_{COG, clock}}$					
$T_{\text{uncertainty}} = \frac{1}{F_{CC}}$	DG clock				
$T_{\text{uncertainty}} = \frac{1}{F_{CC}}$ Where:	 DG_clock				
	 DG_clock Count				
Where:					
Where:	Count				
Where: T Rising Phase Delay	Count COGxPHR				
Where: T Rising Phase Delay Falling Phase Delay	Count COGxPHR COGxPHF				
Where: T Rising Phase Delay Falling Phase Delay Rising dead band	COGxPHR COGxPHF COGxDBR				

EXAMPLE 27-1: TIMER UNCERTAINTY

Given: Count = Ah = 10d $F_{COCC} = C = 1 = 8MHz$

$$COG_Clock = 8MHZ$$

Therefore:

$$T_{\text{uncertainty}} = \frac{1}{F_{COG_clock}}$$

$$= \frac{1}{8MHz} = 125ns$$

Proof:

$$T_{\min} = \frac{Count}{F_{COG_clock}}$$
$$= 125ns \bullet 10d = 1.25 \mu s$$

$$T_{\max} = \frac{Count + 1}{F_{COG_clock}}$$

$$= 125ns \bullet (10d+1)$$

 $= 1.375 \mu s$

Therefore:

$$T_{\text{uncertainty}} = T_{\text{max}} - T_{\text{min}}$$
$$= 1.375 \,\mu s - 1.25 \,\mu s$$
$$= 125 n s$$

27.10 Auto-Shutdown Control

Auto-shutdown is a method to immediately override the COG output levels with specific overrides that allow for safe shutdown of the circuit.

The shutdown state can be either cleared automatically or held until cleared by software. In either case, the shutdown overrides remain in effect until the first rising event after the shutdown is cleared.

27.10.1 SHUTDOWN

The shutdown state can be entered by either of the following two mechanisms:

- Software generated
- External Input

27.10.1.1 Software Generated Shutdown

Setting the ASE bit of the COGxASD0 register (Register 27-11) will force the COG into the shutdown state.

When auto-restart is disabled, the shutdown state will persist until the first rising event after the ASE bit is cleared by software.

When auto-restart is enabled, the ASE bit will clear automatically and resume operation on the first rising event after the shutdown input clears. See Figure 27-15 and Section 27.10.3.2 "Auto-Restart".

27.10.1.2 External Shutdown Source

External shutdown inputs provide the fastest way to safely suspend COG operation in the event of a Fault condition. When any of the selected shutdown inputs go true, the output drive latches are reset and the COG outputs immediately go to the selected override levels without software delay.

Any combination of the input sources can be selected to cause a shutdown condition. Shutdown occurs when the selected source is low. Shutdown input sources include:

- Any input pin selected with the COGxINPPS
- controlComparator 1
- Comparator 1
- Comparator 2Comparator 3
- Comparator 3
 Comparator 4
- CLC2 output/CLC4 output
- Timer2 output/Timer6 output
- Timer4 output/Timer8 output

Shutdown inputs are selected independently with bits of the COGxASD1 register (Register 27-12).

Note:	Shutdown inputs are level sensitive, not edge sensitive. The shutdown state cannot be cleared as long as the shutdown input level persists, except by
	disabling auto-shutdown,

27.10.2 PIN OVERRIDE LEVELS

The levels driven to the output pins, while the shutdown is active, are controlled by the ASDAC<1:0> and ASDBC<1:0> bits of the COGxASD0 register (Register 27-11). ASDAC<1:0> controls the COGxA and COGxC override levels and ASDBC<1:0> controls the COGxB and COGxD override levels. There are four override options for each output pair:

- Forced low
- Forced high
- Tri-state
- PWM inactive state (same state as that caused by a falling event)

Note: The polarity control does not apply to the forced low and high override levels but does apply to the PWM inactive state.

27.10.3 AUTO-SHUTDOWN RESTART

After an auto-shutdown event has occurred, there are two ways to resume operation:

- Software controlled
- Auto-restart

The restart method is selected with the ARSEN bit of the COGxASD0 register. Waveforms of a software controlled automatic restart are shown in Figure 27-15.

27.10.3.1 Software Controlled Restart

When the ARSEN bit of the COGxASD0 register is cleared, software must clear the ASE bit to restart COG operation after an auto-shutdown event.

The COG will resume operation on the first rising event after the ASE bit is cleared. Clearing the shutdown state requires all selected shutdown inputs to be false, otherwise, the ASE bit will remain set.

27.10.3.2 Auto-Restart

When the ARSEN bit of the COGxASD0 register is set, the COG will restart from the auto-shutdown state automatically.

The ASE bit will clear automatically and the COG will resume operation on the first rising event after all selected shutdown inputs go false.

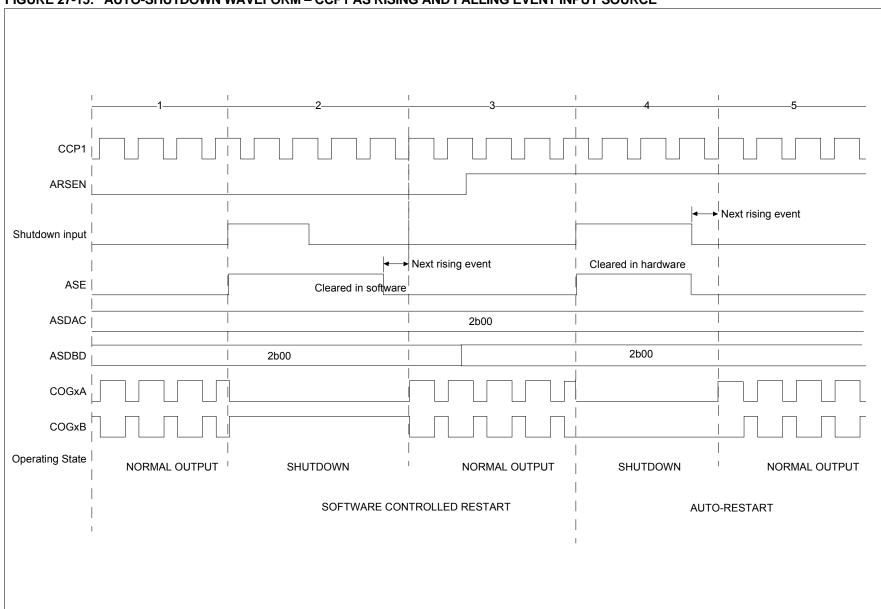


FIGURE 27-15: AUTO-SHUTDOWN WAVEFORM – CCP1 AS RISING AND FALLING EVENT INPUT SOURCE

27.11 Buffer Updates

Changes to the phase, dead-band, and blanking count registers need to occur simultaneously during COG operation to avoid unintended operation that may occur as a result of delays between each register write. This is accomplished with the LD bit of the COGxCON0 register and double buffering of the phase, blanking and dead-band count registers.

Before the COG module is enabled, writing the count registers loads the count buffers without need of the LD bit. However, when the COG is enabled, the count buffer updates are suspended after writing the count registers until after the LD bit is set. When the LD bit is set, the phase, dead-band and blanking register values are transferred to the corresponding buffers synchronous with COG operation. The LD bit is cleared by hardware when the transfer is complete.

27.12 Input and Output Pin Selection

The COG has one selection for an input from a device pin. That one input can be used as rising and falling event source or a fault source. The COGxINPPS register is used to select the pin. Refer to registers xxxPPS (Register 12-1) and RxyPPS (Register 12-2).

The pin PPS control registers are used to enable the COG outputs. Any combination of outputs to pins is possible including multiple pins for the same output. See the RxyPPS control register and Section 12.2 "PPS Outputs" for more details.

27.13 Operation During Sleep

The COG continues to operate in Sleep provided that the COG_clock, rising event, and falling event sources remain active.

The HFINTSOC remains active during Sleep when the COG is enabled and the HFINTOSC is selected as the COG_clock source.

27.14 Configuring the COG

The following steps illustrate how to properly configure the COG to ensure a synchronous start with the rising event input:

- 1. If a pin is to be used for the COG fault or event input, use the COGxINPPS register to configure the desired pin.
- 2. Clear all ANSEL register bits associated with pins that are used for COG functions.
- Ensure that the TRIS control bits corresponding to the COG outputs to be used are set so that all are configured as inputs. The COG module will enable the output drivers as needed later.
- 4. Clear the EN bit, if not already cleared.

- 5. Set desired dead-band times with the COGxDBR and COGxDBF registers and select the source with the RDBS and FDBS bits of the COGxCON1 register.
- 6. Set desired blanking times with the COGxBLKR and COGxBLKF registers.
- 7. Set desired phase delay with the COGxPHR and COGxPHF registers.
- 8. Select the desired shutdown sources with the COGxASD1 register.
- 9. Setup the following controls in COGxASD0 auto-shutdown register:
 - Select both output override controls to the desired levels (this is necessary, even if not using auto-shutdown because start-up will be from a shutdown state).
 - Set the ASE bit and clear the ARSEN bit.
- 10. Select the desired rising and falling event sources with the COGxRIS0, COGxRIS1, COGxFIS0, and COGxFIS1 registers.
- 11. Select the desired rising and falling event modes with the COGxRSIM0, COGxRSIMI1, COGxFSIM0, and COGxFSIM1 registers.
- 12. Configure the following controls in the COGxCON1 register:
 - Set the polarity for each output
 - Select the desired dead-band timing sources
- 13. Configure the following controls in the COGxCON0 register:
 - Set the desired operating mode
 - Select the desired clock source
- 14. If one of the steering modes is selected then configure the following controls in the COGxSTR register:
 - Set the steering bits of the outputs to be used.
 - Set the desired static levels.
- 15. Set the EN bit.
- 16. Set the pin PPS controls to direct the COG outputs to the desired pins.
- 17. If auto-restart is to be used, set the ARSEN bit and the ASE will be cleared automatically. Otherwise, clear the ASE bit to start the COG.

27.15 Register Definitions: COG Control

Long bit name prefixes for the COG peripherals are shown in Table 27-4. Refer to **Section 1.1** "**Register and Bit naming conventions**" for more information **TABLE 27-4**:

Peripheral	Bit Name Prefix
COG1	G1
COG2	G2
COG3	G3
COG4 ⁽¹⁾	G4

Note 1: PIC16(L)F1777/9 only.

REGISTER 27-1: COGxCON0: COG CONTROL REGISTER 0

R/W-0/0	R/W-0/0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
EN	LD	—	CS<	:1:0>		MD<2:0>	
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 7	EN: COGx Enable bit
	1 = Module is enabled
	0 = Module is disabled
bit 6	LD: COGx Load Buffers bit
	 1 = Phase, blanking, and dead-band buffers to be loaded with register values on next input events 0 = Register to buffer transfer is complete
bit 5	Unimplemented: Read as '0'
bit 4-3	CS<1:0>: COGx Clock Selection bits
	11 = Reserved. Do not use.
	10 = COG_clock is HFINTOSC (stays active during Sleep)
	01 = COG_clock is Fosc 00 = COG_clock is Fosc/4
h it 0.0	-
bit 2-0	MD<2:0>: COGx Mode Selection bits
	11x = Reserved. Do not use.
	 101 = COG outputs operate in Push-Pull mode 100 = COG outputs operate in Half-Bridge mode
	011 = COG outputs operate in Reverse Full-Bridge mode
	010 = COG outputs operate in Forward Full-Bridge mode
	001 = COG outputs operate in synchronous steered PWM mode
	000 = COG outputs operate in steered PWM mode

R/W-0/0	R/W-0/0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	
RDBS	FDBS	_		POLD	POLC	POLB	POLA	
bit 7							bit 0	
Legend:								
R = Readable		W = Writable		•	mented bit, read			
u = Bit is unch	anged	x = Bit is unk	nown	-n/n = Value a	at POR and BO	R/Value at all c	other Resets	
1' = Bit is set		'0' = Bit is cle	ared	q = Value dep	pends on condit	ion		
bit 7		· Diaing Evont	Dood bond T	ming Source S	alaat hit			
		•		ming Source S	d timing genera	tion		
					nd timing genera			
bit 6	FDBS: COGx	Falling Event	Dead-band T	iming Source s	elect bit			
	1 = Delay ch							
	$0 = COGx_c$	clock and COG	SxDBF are use	ed for dead-ban	nd timing genera	ation		
bit 5-4	Unimplemen	ted: Read as	'0'					
bit 3	POLD: COG>	D Output Pola	arity Control bi	t				
		vel of COGxD						
	0 = Active le	vel of COGxD	output is high	I				
bit 2	POLC: COG>	C Output Pola	arity Control bi	t				
		vel of COGxC						
		vel of COGxC						
bit 1	POLB: COGxB Output Polarity Control bit 1 = Active level of COGxB output is low							
		evel of COGxB evel of COGxB						
hit O								
bit 0		A Output Pola	•	L				
	1 = Active le		output is IOW					

REGISTER 27-2: COGxCON1: COG CONTROL REGISTER 1

REGISTER 27-3: COGxRIS0: COG RISING EVENT INPUT SELECTION REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	
RIS7	RIS6	RIS5	RIS4	RIS3	RIS2	RIS1	RIS0	
bit 7		•	•	•			bit 0	
Legend:								
R = Readable bit W = Writable bit				U = Unimplemented bit, read as '0'				
u = Bit is unchanged x = Bit is unknown				-n/n = Value at POR and BOR/Value at all other Resets				

q = Value depends on condition

'1' = Bit is set '0' = Bit is cleared

bit 7-0

RIS<7:0>: Source Rising Event Input <n> Source Enable bits⁽¹⁾. See Table 27-5.

1 = Source <n> output is enabled as a rising event input

0 = Source <n> output has no effect on the rising event

Note 1: Any combination of <n> bits can be selected.

REGISTER 27-4: COGxRIS1: COG RISING EVENT INPUT SELECTION REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
RIS15	RIS14	RIS13	RIS12	RIS11	RIS10	RIS9	RIS8
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 15-8

RIS<15:8>: COGx Rising Event Input <n> Source Enable bits⁽¹⁾. See Table 27-5.

1 = Source <n> output is enabled as a rising event input

0 = Source <n> output has no effect on the rising event

Note 1: Any combination of <n> bits can be selected.

TABLE 27-5: RISING/FALLING EVENT INPUT SOURCES

Bit <n></n>	COG1	COG2	COG3 ⁽¹⁾	COG3 ⁽²⁾	COG4 ⁽¹⁾
15	LC4_out	LC4_out	LC4_out	LC4_out	LC4_out
14	LC3_out	LC3_out	LC3_out	LC3_out	LC3_out
13	LC2_out	LC2_out	LC2_out	LC2_out	LC2_out
12	LC1_out	LC1_out	LC1_out	LC1_out	LC1_out
11	MD1_out	MD2_out	MD3_out	MD3_out	MD4_out
10	PWM6_output	PWM6_output	PWM12_output	Reserved	PWM12_output
9	PWM5_output	PWM5_output	PWM11_output	PWM11_output	PWM11_output
8	PWM4_output	PWM4_output	PWM10_output	Reserved	PWM10_output
7	PWM3_output	PWM3_output	PWM9_output	PWM9_output	PWM9_output
6	CCP2_out	CCP2_out	CCP8_out	CCP7_out	CCP8_out
5	CCP1_out	CCP1_out	CCP7_out	CCP1_out	CCP7_out
4	sync_CM4_out	sync_CM4_out	sync_CM8_out	sync_CM6_out	sync_CM8_out
3	sync_CM3_out	sync_CM3_out	sync_CM7_out	sync_CM5_out	sync_CM7_out
2	sync_CM2_out	sync_CM2_out	sync_CM6_out	sync_CM2_out	sync_CM6_out
1	sync_CM1_out	sync_CM1_out	sync_CM5_out	sync_CM1_out	sync_CM5_out
0	Pin selected with COG1PPS	Pin selected with COG2PPS	Pin selected with COG3PPS	Pin selected with COG3PPS	Pin selected with COG4PPS

Note 1: PIC16(L)F1777/9 only.

2: PIC16(L)F1778 only.

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
RSIM7	RSIM6	RSIM5	RSIM4	RSIM3	RSIM2	RSIM1	RSIM0
bit 7							bit 0
							Dit
Legend:							

U = Unimplemented bit, read as '0'

-n/n = Value at POR and BOR/Value at all other Resets

REGISTER 27-5: COGxRSIM0: COG RISING EVENT SOURCE INPUT MODE REGISTER

 '1' = Bit is set
 '0' = Bit is cleared
 q = Value depends on condition

 bit 7-0
 RSIM<7:0>: Rising Event Input Source <n> Mode bits⁽¹⁾. See Table 27-5.

 RIS<n> = 1:
 1 = Source <n> output low-to-high transition will cause a rising event after rising event phase delay 0 = Source <n> output high level will cause an immediate rising event

RIS<n> = 0:

R = Readable bit

u = Bit is unchanged

Source <n> output has no effect on rising event

W = Writable bit

x = Bit is unknown

Note 1: Any combination of <n> bits can be selected.

REGISTER 27-6: COGxRSIM1: COG RISING EVENT SOURCE INPUT MODE REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
RSIM15	RSIM14	RSIM13	RSIM12	RSIM11	RSIM10	RSIM9	RSIM8
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 15-8 **RSIM<15:8>:** Rising Event Input Source <n> Mode bits⁽¹⁾. See Table 27-5. <u>RIS<n> = 1:</u> 1 = Source <n> output low-to-high transition will cause a rising event after rising event phase delay 0 = Source <n> output high level will cause an immediate rising event <u>RIS<n> = 0:</u> Source <n> output has no effect on rising event

Note 1: Any combination of <n> bits can be selected.

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
FIS7	FIS6	FIS5	FIS4	FIS3	FIS2	FIS1	FIS0
bit 7							bit 0
Legend:							
R = Readable b	oit	W = Writable I	bit	U = Unimpler	mented bit, read	as '0'	
u = Bit is uncha	anged	x = Bit is unkn	iown	-n/n = Value a	at POR and BOI	R/Value at all o	ther Resets
'1' = Bit is set		'0' = Bit is clea	ared	q = Value dep	pends on condit	ion	

REGISTER 27-7: COGxFIS0: COG FALLING EVENT INPUT SELECTION REGISTER

bit 7-0 **FIS<7:0>:** Falling Event Input Source <n> Enable bits⁽¹⁾. See Table 27-5.

1 = Source <n> output is enabled as a falling event input

0 = Source <n> output has no effect on the falling event

Note 1: Any combination of <n> bits can be selected.

REGISTER 27-8: COGxFIS1: COG FALLING EVENT INPUT SELECTION REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
FIS15	FIS14	FIS13	FIS12	FIS11	FIS10	FIS9	FIS8
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 15-8 **FIS<15:8>:** Falling Event Input Source <n> Enable bits⁽¹⁾. See Table 27-5.

1 = Source <n> output is enabled as a falling event input

0 = Source <n> output has no effect on the falling event

Note 1: Any combination of <n> bits can be selected.

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	
FSIM7	FSIM6	FSIM5	FSIM4	FSIM3	FSIM2	FSIM1	FSIM0	
bit 7							bit 0	
Legend:								
R = Readable	bit	W = Writable	bit	U = Unimpler	nented bit, read	as '0'		
u = Bit is unchanged x = Bit is unknown				-n/n = Value at POR and BOR/Value at all other Resets				
'1' = Bit is set '0' = Bit is cleared q = Value depends on condition								

REGISTER 27-9: COGxFSIM0: COG FALLING EVENT SOURCE INPUT MODE REGISTER

bit 7-0 FSIM<7:0>: Falling Event Input Source <n> Mode bits⁽¹⁾. See Table 27-5. FIS<n> = 1: 1 = Source <n> output high-to-low transition will cause a falling event after falling event phase delay 0 = Source <n> output low level will cause an immediate falling event FIS<n> = 0: Source <n> output has no effect on falling event

Note 1: Any combination of <n> bits can be selected.

REGISTER 27-10: COGxFSIM1: COG FALLING EVENT SOURCE INPUT MODE REGISTER

	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
FSIM15	FSIM14	FSIM13	FSIM12	FSIM11	FSIM10	FSIM9	FSIM8
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 15-8 FSIM<15:8>: Falling Event Input Source <n> Mode bits⁽¹⁾. See Table 27-5. FIS<n> = 1: 1 = Source <n> output high-to-low transition will cause a falling event after falling event phase delay 0 = Source <n> output low level will cause an immediate falling event FIS<n> = 0: Source <n> output has no effect on falling event

Note 1: Any combination of <n> bits can be selected.

R/W-0/0	R/W-0/0	R/W-0/0	R/W-1/1	R/W-0/0	R/W-1/1	U-0	U-0			
ASE	ARSEN	ASDBD<1:0>		ASDBD<1:0> ASDAC<1:0>			_			
bit 7	·	·		÷			bit 0			
Legend:										
R = Readable	e bit	W = Writable	bit	U = Unimplem	ented bit, read a	as '0'				
u = Bit is unchanged x = Bit is unknown -n/n = Value at POR ar						/Value at all oth	ner Resets			
'1' = Bit is se	t	'0' = Bit is clea	ared	q = Value depe	ends on conditio	n				
bit 7	ASE: Auto-S	hutdown Event	Status bit							
		n the shutdown								
	0 = COG is e	either not in the	shutdown sta	ate or will exit the	e shutdown state	e on the next ris	sing event			
bit 6	ARSEN: Auto	o-Restart Enabl	e bit							
		tart is enabled								
	0 = Auto-res	tart is disabled								
bit 5-4	ASDBD<1:0:	COGxB and	COGxD Auto	-shutdown Overi	ide Level Selec	t bits				
	•	•		COGxD when s						
	 10 = A logic '0' is placed on COGxB and COGxD when shutdown is active 01 = COGxB and COGxD are tri-stated when shutdown is active 									
							han abutdawa			
	is activ		ie pin, includi	ng polarity, is pla			nen shuluowi			
bit 3-2	ASDAC<1:0:	-: COGxA and	COGxC Auto	-shutdown Overi	ide Level Select	t bits				
	11 = A logic	'1' is placed on	COGxA and	COGxC when s	hutdown is activ	'e				
		 11 = A logic '1' is placed on COGxA and COGxC when shutdown is active 10 = A logic '0' is placed on COGxA and COGxC when shutdown is active 								
				vhen shutdown i						
	00 = The ina is active		ie pin, includir	ng polarity, is pla	ced on COGxA	and COGxC wl	hen shutdowr			
bit 1-0	Unimplemer	ted: Read as '	0'							

REGISTER 27-11: COGxASD0: COG AUTO-SHUTDOWN CONTROL REGISTER 0

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
AS7E	AS6E	AS5E	AS4E	AS3E	AS2E	AS1E	AS0E
bit 7 b							bit 0
Legend:							
R = Readable	R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'			as '0'			
u = Bit is unc	hanged	x = Bit is unkr	nown	-n/n = Value at POR and BOR/Value at all other Resets			

q = Value depends on condition

REGISTER 27-12: COGxASD1: COG AUTO-SHUTDOWN CONTROL REGISTER 1

bit 7-0 **AS<7:0>E:** Auto-shutdown Source <n> Enable bits⁽¹⁾. See Table 27-6.

1 = COGx is shutdown when source <n> output is low

0 = Source <n> has no effect on shutdown

'0' = Bit is cleared

Note 1: Any combination of <n> bits can be selected.

Bit <n></n>	COG1	COG2	COG3 ⁽²⁾	COG3 ⁽³⁾	COG4 ⁽²⁾
7	TMR4_postscaled ⁽¹⁾	TMR4_postscaled ⁽¹⁾	TMR8_postscaled ⁽¹⁾	TMR8_postscaled ⁽¹⁾	TMR8_postscaled ⁽¹⁾
6	TMR2_postscaled ⁽¹⁾	TMR2_postscaled ⁽¹⁾	TMR6_postscaled ⁽¹⁾	TMR6_postscaled ⁽¹⁾	TMR6_postscaled ⁽¹⁾
5	LC2_out	LC2_out	LC4_out	LC4_out	LC4_out
4	sync_CM4_out	sync_CM4_out	sync_CM8_out	sync_CM6_out	sync_CM8_out
3	sync_CM3_out	sync_CM3_out	sync_CM7_out	sync_CM5_out	sync_CM7_out
2	sync_CM2_out	sync_CM2_out	sync_CM6_out	sync_CM2_out	sync_CM6_out
1	sync_CM1_out	sync_CM1_out	sync_CM5_out	sync_CM1_out	sync_CM5_out
0	Pin selected by COG1PPS	Pin selected by COG2PPS	Pin selected by COG3PPS	Pin selected by COG3PPS	Pin selected by COG4PPS

Note 1: Shutdown when source is high.

2: PIC16(L)F1777/9 only.

'1' = Bit is set

3: PIC16(L)F1778 only.

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
SDATD	SDATC	SDATB	SDATA	STRD	STRC	STRB	STRA
bit 7							bit
Legend:	la hit		L:+		antad bit used	(O'	
R = Readabl		W = Writable		•	ented bit, read		hor Dogota
u = Bit is unchangedx = Bit is unknown-n/n = Value at POR and BOR/Value at all other R'1' = Bit is set'0' = Bit is clearedq = Value depends on condition							
I = DILIS SE	÷l		areu	q = value depe		011	
bit 7	SDATD: CO	GxD Static Out	out Data bit				
		static data is h					
		static data is lo	•				
bit 6	SDATC: CO	GxC Static Out	put Data bit				
		static data is h	•				
L:1. F		static data is lo					
bit 5		GxB Static Out					
		static data is lo	0				
bit 4		GxA Static Outp					
		static data is hi					
	0 = COGxA	static data is lo	Ŵ				
bit 3		SxD Steering Co					
				form with polarity		POLD bit	
bit 2		SxC Steering Co		determined by th	he SDATD bit		
		0		form with polarity	control from F		
				determined by th		OLO DI	
bit 1	STRB: COG	SxB Steering Co	ntrol bit				
				orm with polarity		OLB bit	
		-		determined by th	ne SDATB bit		
bit 0		SxA Steering Co					
	1 = COGxA 0 = COGxA			form with polarity		OLA bit	

REGISTER 27-13: COGxSTR: COG STEERING CONTROL REGISTER 1⁽¹⁾

U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	
				DBR	<5:0>			
bit 7		·					bit 0	
Legend:								
R = Readable	R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'				l as '0'			
u = Bit is unch	anged	x = Bit is unknown		-n/n = Value at POR and BOR/Value at all other Rese				
'1' = Bit is set	' = Bit is set '0' = Bit is cleared		ared	q = Value depends on condition				

bit 7-6	Unimplemented: Read as '0'
bit 5-0	DBR<5:0>: Rising Event Dead-Band Count Value bits
	RDBS = 1:= Number of delay chain element periods to delay primary output after rising eventRDBS = 0:= Number of COGx clock periods to delay primary output after rising event

REGISTER 27-15: COGxDBF: COG FALLING EVENT DEAD-BAND COUNT REGISTER

U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
		DBF<5:0>					
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 7-6 Unimplemented: Read as '0'

DBF<5:0>: Falling Event Dead-Band Count Value bits

FDBS = 1:

bit 5-0

= Number of delay chain element periods to delay complementary output after falling event input FDBS = 0:

= Number of COGx clock periods to delay complementary output after falling event input

REGISTER 27-16: COGxBLKR: COG RISING EVENT BLANKING COUNT REGISTER

U-0	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
			BLKF	R<5:0>		
	·					bit 0
bit	W = Writable	/ = Writable bit U = Unimplemented bit, read as '0'				
anged	x = Bit is unkr	it is unknown -n/n = Value at POR and BOR/Value at all other Rese				
	'0' = Bit is clea	cleared q = Value depends on condition				
	U-0	bit W = Writable anged x = Bit is unkr	bit W = Writable bit	bit W = Writable bit U = Unimplen anged x = Bit is unknown -n/n = Value a	BLKR<5:0> bit W = Writable bit U = Unimplemented bit, read anged x = Bit is unknown -n/n = Value at POR and BO	BLKR < 5:0 > bit W = Writable bit U = Unimplemented bit, read as '0' anged x = Bit is unknown -n/n = Value at POR and BOR/Value at all or

bit 7-6 Unimplemented: Read as '0'

bit 5-0 BLKR<5:0>: Rising Event Blanking Count Value bits

= Number of COGx clock periods to inhibit falling event inputs

REGISTER 27-17: COGxBLKF: COG FALLING EVENT BLANKING COUNT REGISTER

U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0		
			BLKF<5:0>						
bit 7							bit 0		

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 7-6 Unimplemented: Read as '0'

bit 5-0 BLKF<5:0>: Falling Event Blanking Count Value bits

= Number of COGx clock periods to inhibit rising event inputs

REGISTER 27-18: COGxPHR: COG RISING EVENT PHASE DELAY COUNT REGISTER

U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	_			PHR	<5:0>		
bit 7							bit 0
Legend:							
R = Readable b	oit	W = Writable I	oit	U = Unimplem	nented bit, read	as '0'	
u = Bit is unchanged x = Bit is unknown			-n/n = Value at POR and BOR/Value at all other Resets				
'1' = Bit is set		'0' = Bit is cleared q = Value depends on condition					

bit 7-6	Unimplemented: Read as '0'
---------	----------------------------

bit 5-0 PHR<5:0>: Rising Event Phase Delay Count Value bits

= Number of COGx clock periods to delay rising event

REGISTER 27-19: COGxPHF: COG FALLING EVENT PHASE DELAY COUNT REGISTER

U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	
—	_		PHF<5:0>					
bit 7							bit 0	

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 7-6 Unimplemented: Read as '0'

bit 5-0 PHF<5:0>: Falling Event Phase Delay Count Value bits

= Number of COGx clock periods to delay falling event

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ANSELA	_	_	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0	177
ANSELB	_	_	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	182
ANSELC	ANSC7	ANSC6	ANSC5	ANSC4	ANSC3	ANSC2	—	—	187
COGxASD0	ASE	ARSEN	ASDB	D<1:0>	ASDA	C<1:0>		—	384
COGxASD1	AS7E	AS6E	AS5E	AS4E	AS3E	AS2E	AS1E	AS0E	385
COGxBLKR	_	_		•	BLKR	<5:0>		•	388
COGxBLKF	—	_			BLKF	<5:0>			388
COGxCON0	EN	LD	_	CS<	:1:0>		MD<2:0>		378
COGxCON1	RDBS	FDBS	_	_	POLD	POLC	POLB	POLA	379
COGxDBR		_		•	DBR	<5:0>		•	387
COGxDBF	_	_		DBF<5:0>					
COGxFIS0	FIS7	FIS6	FIS5	FIS4	FIS3	FIS2	FIS1	FIS0	382
COGxFIS1	FIS15	FIS14	FIS13	FIS12	FIS11	FIS10	FIS9	FIS8	382
COGxFSIM0	FSIM7	FSIM6	FSIM5	FSIM4	FSIM3	FSIM2	FSIM1	FSIM0	383
COGxFSIM1	FSIM15	FSIM14	FSIM13	FSIM12	FSIM11	FSIM10	FSIM9	FSIM8	383
COGxPHR	—	—		•	PHR	<5:0>		•	389
COGxPHF	_	_			PHF	<5:0>			389
COGxPPS	_	_			COG1P	PS<5:0>			205, 207
COGxRIS0	RIS7	RIS6	RIS5	RIS4	RIS3	RIS2	RIS1	RIS0	380
COGxRIS1	RIS15	RIS14	RIS13	RIS12	RIS11	RIS10	RIS9	RIS8	380
COGxRSIM0	RSIM7	RSIM6	RSIM5	RSIM4	RSIM3	RSIM2	RSIM1	RSIM0	381
COGxRSIM1	RSIM15	RSIM14	RSIM13	RSIM12	RSIM11	RSIM10	RSIM9	RSIM8	381
COGxSTR	SDATD	SDATC	SDATB	SDATA	STRD	STRC	STRB	STRA	386
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	132
RxyPPS	_	_			RxyPP	S<5:0>			205

TABLE 27-7: SUMMARY OF REGISTERS ASSOCIATED WITH COGx⁽¹⁾

Legend: x = unknown, u = unchanged, – = unimplemented locations read as '0'. Shaded cells are not used by COG. **Note 1:** COG4 is available on PIC16(L)F1777/9 only.

28.0 CONFIGURABLE LOGIC CELL (CLC)

The Configurable Logic Cell (CLCx) provides programmable logic that operates outside the speed limitations of software execution. The logic cell takes up to 32 input signals and, through the use of configurable gates, reduces the 32 inputs to four logic lines that drive one of eight selectable single-output logic functions.

Input sources are a combination of the following:

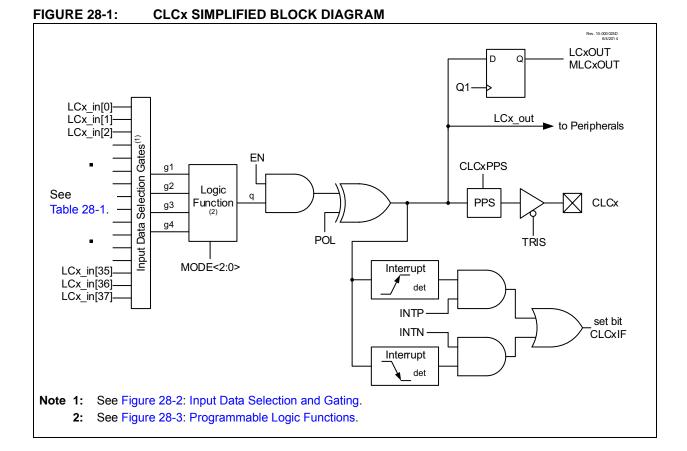
- · I/O pins
- Internal clocks
- · Peripherals
- · Register bits

The output can be directed internally to peripherals and to an output pin.

Refer to Figure 28-1 for a simplified diagram showing signal flow through the CLCx.

Possible configurations include:

- Combinatorial Logic
 - AND
 - NAND
 - AND-OR
 - AND-OR-INVERT
 - OR-XOR
 - OR-XNOR
- Latches
 - S-R
 - Clocked D with Set and Reset
 - Transparent D with Set and Reset
 - Clocked J-K with Reset



28.1 CLCx Setup

Programming the CLCx module is performed by configuring the four stages in the logic signal flow. The four stages are:

- · Data selection
- Data gating
- · Logic function selection
- · Output polarity

Each stage is setup at run time by writing to the corresponding CLCx Special Function Registers. This has the added advantage of permitting logic reconfiguration on-the-fly during program execution.

28.1.1 DATA SELECTION

There are 32 signals available as inputs to the configurable logic. Four 32-input multiplexers are used to select the inputs to pass on to the next stage.

Data selection is through four multiplexers as indicated on the left side of Figure 28-2. Data inputs in the figure are identified by a generic numbered input name.

Table 28-1 correlates the generic input name to the actual signal for each CLC module. The column labeled dy indicates the MUX selection code for the selected data input. DxS is an abbreviation for the MUX select input codes: D1S<4:0> through D4S<4:0>.

Data inputs are selected with the CLCxSEL0 through CLCxSEL3 registers (Register 28-3 through Register 28-6).

Note: Data selections are undefined at power-up.

TABLE 28-1: CLCx DATA INPUT SELECTION

Data Input	dy DxS	CLCx
LCx_in[54]	110110	MD1_out or MD2_out or MD3_out
LCx_in[53]	110101	FOSC
LCx_in[52]	110100	HFINTOSC
LCx_in[51]	110011	LFINTOSC
LCx_in[50]	110010	FRC (ADC RC clock)
LCx_in[49]	110001	IOCIF set
LCx_in[48]	110000	Timer8_postscaled
LCx_in[47]	101111	Timer6_postscaled
LCx_in[46]	101110	Timer4_postscaled
LCx_in[45]	101101	Timer2_postscaled
LCx_in[44]	101100	Timer5 overflow
LCx_in[43]	101011	Timer3 overflow
LCx_in[42]	101010	Timer1 overflow
LCx_in[41]	101001	Timer0 overflow
LCx_in[40]	101000	EUSART RX
LCx_in[39]	100111	EUSART TX
LCx_in[38]	100110	ZCD1_output
LCx_in[37]	100101	MSSP1 SDO/SDA
LCx_in[36]	100100	MSSP1 SCL/SCK

TABLE 28-1: CLCx DATA INPUT SELECTION

Data Input	dy DxS	CLCx		
LCx_in[35]	100011	PWM12_out ⁽¹⁾		
LCx_in[34]	100010	PWM11_out		
LCx_in[33]	100001	PWM6_out		
LCx_in[32]	100000	PWM5_out		
LCx_in[31]	011111	PWM10_out ⁽¹⁾		
LCx_in[30]	011110	PWM9_out		
LCx_in[29]	011101	PWM4_out		
LCx_in[28]	011100	PWM3_out		
LCx_in[27]	011011	CCP8_out ⁽¹⁾		
LCx_in[26]	011010	CCP7_out		
LCx_in[25]	011001	CCP2_out		
LCx_in[24]	011000	CCP1_out		
LCx_in[23]	010111	COG4B ⁽¹⁾		
LCx_in[22]	010110	COG4A ⁽¹⁾		
LCx_in[21]	010101	COG3B		
LCx_in[20]	010100	COG3A		
LCx_in[19]	010011	COG2B		
LCx_in[18]	010010	COG2A		
LCx_in[17]	010001	COG1B		
LCx_in[16]	010000	COG1A		
LCx_in[15]	001111	sync_C8OUT ⁽¹⁾		
LCx_in[14]	001110	sync_C7OUT ⁽¹⁾		
LCx_in[13]	001101	sync_C6OUT		
LCx_in[12]	001100	sync_C5OUT		
LCx_in[11]	001011	sync_C4OUT		
LCx_in[10]	001010	sync_C3OUT		
LCx_in[9]	001001	sync_C2OUT		
LCx_in[8]	001000	sync_C1OUT		
LCx_in[7]	000111	LC4_out from the CLC4		
LCx_in[6]	000110	LC3_out from the CLC3		
LCx_in[5]	000101	LC2_out from the CLC2		
LCx_in[4]	000100	LC1_out from the CLC1		
LCx_in[3]	000011	CLCIN3 pin input selected in CLCIN3PPS register		
LCx_in[2]	000010	CLCIN2 pin input selected in CLCIN2PPS register		
LCx_in[1]	000001	CLCIN1 pin input selected in CLCIN1PPS register		
LCx_in[0]	000000	CLCIN0 pin input selected in CLCIN0PPS register		

Note 1: PIC16(L)F1777/9 only.

28.1.2 DATA GATING

Outputs from the input multiplexers are directed to the desired logic function input through the data gating stage. Each data gate can direct any combination of the four selected inputs.

Note: Data gating is undefined at power-up.

The gate stage is more than just signal direction. The gate can be configured to direct each input signal as inverted or non-inverted data. Directed signals are ANDed together in each gate. The output of each gate can be inverted before going on to the logic function stage.

The gating is in essence a 1-to-4 input AND/NAND/OR/NOR gate. When every input is inverted and the output is inverted, the gate is an OR of all enabled data inputs. When the inputs and output are not inverted, the gate is an AND or all enabled inputs.

Table 28-2 summarizes the basic logic that can be obtained in gate 1 by using the gate logic select bits. The table shows the logic of four input variables, but each gate can be configured to use less than four. If no inputs are selected, the output will be zero or one, depending on the gate output polarity bit.

TABLE 28-2: DATA GATING LOGIC

CLCxGLS0	G1POL	Gate Logic
0x55	1	AND
0x55	0	NAND
0xAA	1	NOR
0xAA	0	OR
0x00	0	Logic 0
0x00	1	Logic 1

It is possible (but not recommended) to select both the true and negated values of an input. When this is done, the gate output is zero, regardless of the other inputs, but may emit logic glitches (transient-induced pulses). If the output of the channel must be zero or one, the recommended method is to set all gate bits to zero and use the gate polarity bit to set the desired level.

Data gating is configured with the logic gate select registers as follows:

- Gate 1: CLCxGLS0 (Register 28-7)
- Gate 2: CLCxGLS1 (Register 28-8)
- Gate 3: CLCxGLS2 (Register 28-9)
- Gate 4: CLCxGLS3 (Register 28-10)

Register number suffixes are different than the gate numbers because other variations of this module have multiple gate selections in the same register. Data gating is indicated in the right side of Figure 28-2. Only one gate is shown in detail. The remaining three gates are configured identically with the exception that the data enables correspond to the enables for that gate.

28.1.3 LOGIC FUNCTION

There are eight available logic functions including:

- AND-OR
- OR-XOR
- AND
- S-R Latch
- D Flip-Flop with Set and Reset
- D Flip-Flop with Reset
- J-K Flip-Flop with Reset
- · Transparent Latch with Set and Reset

Logic functions are shown in Figure 28-3. Each logic function has four inputs and one output. The four inputs are the four data gate outputs of the previous stage. The output is fed to the inversion stage and from there to other peripherals, an output pin, and back to the CLCx itself.

28.1.4 OUTPUT POLARITY

The last stage in the Configurable Logic Cell is the output polarity. Setting the POL bit of the CLCxCON register inverts the output signal from the logic stage. Changing the polarity while the interrupts are enabled will cause an interrupt for the resulting output transition.

28.1.5 CLCx SETUP STEPS

The following steps should be followed when setting up the CLCx:

- Disable CLCx by clearing the EN bit.
- Select desired inputs using CLCxSEL0 through CLCxSEL3 registers (See Table 28-1).
- · Clear any associated ANSEL bits.
- Set all TRIS bits associated with inputs.
- · Clear all TRIS bits associated with outputs.
- Enable the chosen inputs through the four gates using CLCxGLS0, CLCxGLS1, CLCxGLS2, and CLCxGLS3 registers.
- Select the gate output polarities with the POLy bits of the CLCxPOL register.
- Select the desired logic function with the MODE<2:0> bits of the CLCxCON register.
- Select the desired polarity of the logic output with the POL bit of the CLCxPOL register. (This step may be combined with the previous gate output polarity step).
- If driving a device pin, set the desired pin PPS control register and also clear the TRIS bit corresponding to that output.
- If interrupts are desired, configure the following bits:
 - Set the INTP bit in the CLCxCON register for rising event.
 - Set the INTN bit in the CLCxCON register for falling event.
 - Set the CLCxIE bit of the associated PIE registers.
 - Set the GIE and PEIE bits of the INTCON register.
- Enable the CLCx by setting the EN bit of the CLCxCON register.

28.2 CLCx Interrupts

An interrupt will be generated upon a change in the output value of the CLCx when the appropriate interrupt enables are set. A rising edge detector and a falling edge detector are present in each CLC for this purpose.

The CLCxIF bit of the associated PIR registers will be set when either edge detector is triggered and its associated enable bit is set. The INTP enables rising edge interrupts and the INTN bit enables falling edge interrupts. Both are located in the CLCxCON register.

To fully enable the interrupt, set the following bits:

- · EN bit of the CLCxCON register
- · CLCxIE bit of the associated PIE registers
- INTP bit of the CLCxCON register (for a rising edge detection)
- INTN bit of the CLCxCON register (for a falling edge detection)
- · PEIE and GIE bits of the INTCON register

The CLCxIF bit of the associated PIR registers, must be cleared in software as part of the interrupt service. If another edge is detected while this flag is being cleared, the flag will still be set at the end of the sequence.

28.3 Output Mirror Copies

Mirror copies of all LCxCON output bits are contained in the CLCxDATA register. Reading this register reads the outputs of all CLCs simultaneously. This prevents any reading skew introduced by testing or reading the CLCxOUT bits in the individual CLCxCON registers.

28.4 Effects of a Reset

The CLCxCON register is cleared to zero as the result of a Reset. All other selection and gating values remain unchanged.

28.5 Operation During Sleep

The CLC module operates independently from the system clock and will continue to run during Sleep, provided that the input sources selected remain active.

The HFINTOSC remains active during Sleep when the CLC module is enabled and the HFINTOSC is selected as an input source, regardless of the system clock source selected.

In other words, if the HFINTOSC is simultaneously selected as the system clock and as a CLC input source, when the CLC is enabled, the CPU will go idle during Sleep, but the CLC will continue to operate and the HFINTOSC will remain active.

This will have a direct effect on the Sleep mode current.

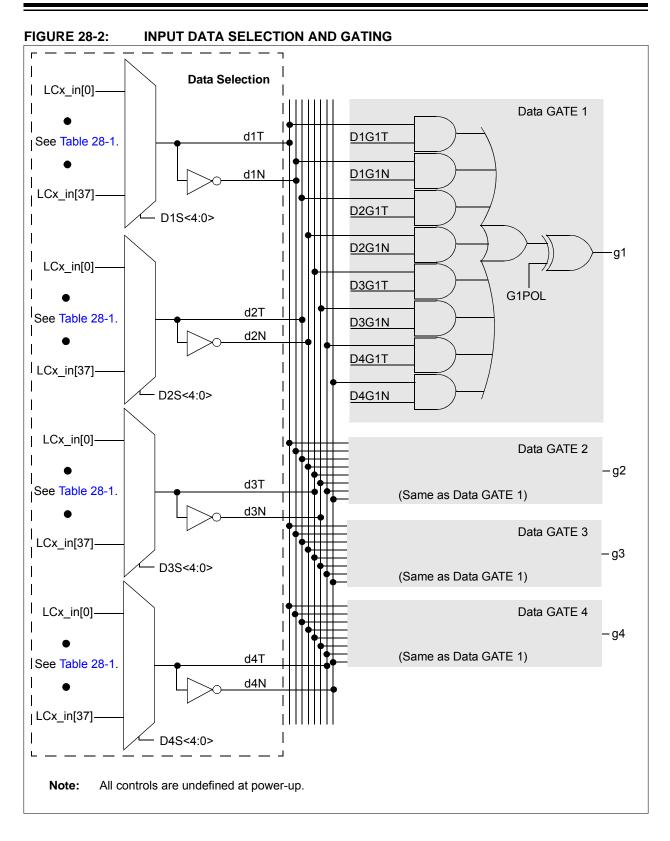
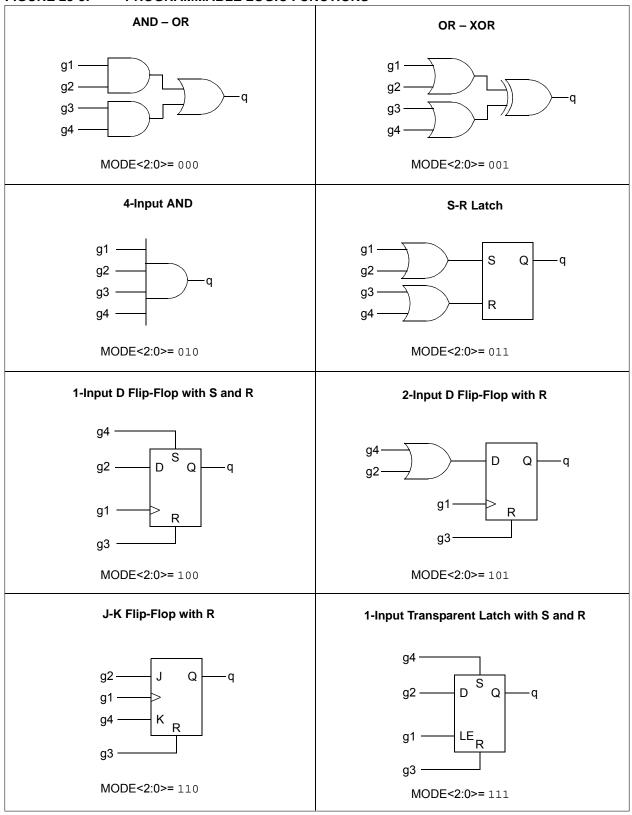


FIGURE 28-3: PROGRAMMABLE LOGIC FUNCTIONS



28.6 Register Definitions: CLC Control

Long bit name prefixes for the CLC peripherals are shown in Table 28-3. Refer to **Section 1.1** "**Register and Bit naming conventions**" for more information **TABLE 28-3**:

Peripheral	Bit Name Prefix
CLC1	LC1
CLC2	LC2
CLC3	LC3
CLC4	LC4

REGISTER 28-1: CLCxCON: CONFIGURABLE LOGIC CELL CONTROL REGISTER

R/W-0/0	U-0	R-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
EN	—	OUT	INTP	INTN	MODE<2:0>		
bit 7							bit 0

Legend:				
R = Readable bit		W = Writable bit	U = Unimplemented bit, read as '0'	
u = Bit is unchanged		x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets	
'1' = Bit is set		'0' = Bit is cleared		
bit 7	•	able Logic Cell Enable bi		
		able logic cell is enabled		
	0 = Configur	able logic cell is disabled	and has logic zero output	
bit 6	Unimplemer	nted: Read as '0'		
bit 5 OUT: Configurable Logic Cell Data Output bit				
	Read-only: lo	ogic cell output data, after	POL; sampled from lcx_out wire.	
bit 4	INTP: Config	urable Logic Cell Positive	e Edge Going Interrupt Enable bit	
	1 = CLCxIF	will be set when a rising e	edge occurs on lcx_out	
	0 = CLCxIF	will not be set		
bit 3	INTN: Config	urable Logic Cell Negativ	e Edge Going Interrupt Enable bit	
	1 = CLCxIF	will be set when a falling	edge occurs on lcx_out	
	0 = CLCxIF	will not be set		
bit 2-0	MODE<2:0>:	: Configurable Logic Cell	Functional Mode bits	
	111 = Cell is	3 1-input transparent latch	n with S and R	
		3 J-K flip-flop with R		
		2-input D flip-flop with R		
		S 1-input D flip-flop with S	and R	
	011 = Cell is	s S-R latch s 4-input AND		
	010 = Cell is			
	000 = Cell is			
		-		

R/W-0/0	U-0	U-0	U-0	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
POL	—	_	—	G4POL	G3POL	G2POL	G1POL
bit 7							bit 0
Lonondi							
Legend:	. 1. 11		L *1				
R = Readable		W = Writable			mented bit, read		
u = Bit is und	•	x = Bit is unkr	nown	-n/n = Value	at POR and BO	R/Value at all c	other Resets
'1' = Bit is se	t	'0' = Bit is clea	ared				
bit 7		Polarity Control					
		ut of the logic of					
	•	ut of the logic of		erted			
bit 6-4	Unimplemen	ted: Read as '	0'				
bit 3	G4POL: Gate	e 4 Output Pola	rity Control b	bit			
	•	ut of gate 4 is i ut of gate 4 is r		n applied to the	logic cell		
bit 2	-	e 3 Output Pola		oit			
		ut of gate 3 is i ut of gate 3 is r		n applied to the	logic cell		
bit 1	G2POL: Gate	e 2 Output Pola	rity Control b	oit			
	 1 = The output of gate 2 is inverted when applied to the logic cell 0 = The output of gate 2 is not inverted 						
bit 0	G1POL: Gate	e 1 Output Pola	rity Control b	oit			
	1 = The outp	•	nverted wher	n applied to the	logic cell		

REGISTER 28-2: CLCxPOL: SIGNAL POLARITY CONTROL REGISTER

REGISTER 28-3: CLCxSEL0: GENERIC CLCx DATA 1 SELECT REGISTER

U-0	U-0	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
_	_			D1S	<5:0>		
bit 7 bi							bit 0
Legend:							
R = Readable bit W = Writable bit		bit	U = Unimplemented bit, read as '0'				
u = Bit is unchanged x = Bit is unknown -n/n = Value at POR and BOR/Value at a				R/Value at all o	ther Resets		

bit 7-6	Unimplemented: Read as '0'

'1' = Bit is set

bit 5-0	D1S<5:0>: CLCx Data1 Input Selection bits
	See Table 28-1.

REGISTER 28-4: CLCxSEL1: GENERIC CLCx DATA 2 SELECT REGISTER

'0' = Bit is cleared

U-0	U-0	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
—	_	D2S<5:0>					
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-6 Unimplemented: Read as '0'

bit 5-0 **D2S<5:0>:** CLCx Data 2 Input Selection bits See Table 28-1.

REGISTER 28-5: CLCxSEL2: GENERIC CLCx DATA 3 SELECT REGISTER

U-0	U-0	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
—	—	D3S<5:0>					
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-6 Unimplemented: Read as '0'

bit 5-0 D3S<5:0>: CLCx Data 3 Input Selection bits See Table 28-1.

REGISTER 28-6: CLCxSEL3: GENERIC CLCx DATA 4 SELECT REGISTER

U-0	U-0	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	
—	_		D4S<5:0>					
bit 7							bit 0	
Legend:								
R = Readable I	bit	W = Writable	W = Writable bit U = Unimplemented bit, read as '0'					
u = Bit is uncha	anged	x = Bit is unkr	nown -n/n = Value at POR and BOR/Value at all other Res				ther Resets	
'1' = Bit is set		'0' = Bit is clea	ared					

bit 7-6Unimplemented: Read as '0'bit 5-0D4S<5:0>: CLCx Data 4 Input Selection bits
See Table 28-1.

REGISTER 28-7: CLCxGLS0: GATE 1 LOGIC SELECT REGISTER

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
G1D4T	G1D4N	G1D3T	G1D3N	G1D2T	G1D2N	G1D1T	G1D1N
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7	G1D4T: Gate 1 Data 4 True (non-inverted) bit	
	1 = d4T is gated into g1	
	0 = d4T is not gated into g1	
bit 6	G1D4N: Gate 1 Data 4 Negated (inverted) bit	
	1 = d4N is gated into g1	
	0 = d4N is not gated into g1	
bit 5	G1D3T: Gate 1 Data 3 True (non-inverted) bit	
	1 = d3T is gated into g1	
	0 = d3T is not gated into g1	
bit 4	G1D3N: Gate 1 Data 3 Negated (inverted) bit	
	1 = d3N is gated into g1	
	0 = d3N is not gated into g1	
bit 3	G1D2T: Gate 1 Data 2 True (non-inverted) bit	
	1 = d2T is gated into g1	
	0 = d2T is not gated into g1	
bit 2	G1D2N: Gate 1 Data 2 Negated (inverted) bit	
	1 = d2N is gated into g1	
	0 = d2N is not gated into g1	
bit 1	G1D1T: Gate 1 Data 1 True (non-inverted) bit	
	1 = d1T is gated into g1	
	0 = d1T is not gated into g1	
bit 0	G1D1N: Gate 1 Data 1 Negated (inverted) bit	
	1 = d1N is gated into g1	
	0 = d1N is not gated into g1	

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u		
G2D4T	G2D4N	G2D3T	G2D3N	G2D2T	G2D2N	G2D1T	G2D1N		
bit 7							bit C		
Levende									
Legend: R = Readable	hit	W = Writable	hit		nented bit, read	ac 'O'			
u = Bit is unch		x = Bit is unkr		•	at POR and BO		thar Pacata		
(1) = Bit is unch(1) = Bit is set	angeu	0' = Bit is cle			at FOR and BO	N value at all 0			
I = DILIS SEL			areu						
bit 7	G2D4T: Gate	e 2 Data 4 True	(non-inverted)) bit					
	1 = d4T is g		,						
	0 = d4T is n	ot gated into g2							
bit 6	G2D4N: Gat	e 2 Data 4 Neg	ated (inverted)	bit					
	1 = d4N is g								
		ot gated into g2							
bit 5		e 2 Data 3 True	(non-inverted)) bit					
	1 = d3T is g	ated into g2 ot gated into g2							
bit 4		e 2 Data 3 Neg		bit					
	1 = d3N is g	-		bit					
		ot gated into g2	2						
bit 3	G2D2T: Gate	e 2 Data 2 True	(non-inverted)) bit					
	1 = d2T is g	ated into g2							
	0 = d2T is n	ot gated into g2							
bit 2		e 2 Data 2 Neg	ated (inverted)	bit					
	1 = d2N is g		,						
L :1 1		ot gated into g2		. h.:+					
bit 1		T: Gate 2 Data 1 True (non-inverted) bit 1T is gated into g2							
		ot gated into g2							
bit 0		e 2 Data 1 Neg		bit					
-	1 = d1N is g	•							
	0 = d1N is n								

REGISTER 28-8: CLCxGLS1: GATE 2 LOGIC SELECT REGISTER

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u				
G3D4T	G3D4N	G3D3T	G3D3N	G3D2T	G3D2N	G3D1T	G3D1N				
bit 7	·						bit (
Legend:											
R = Readable	e bit	W = Writable	bit	U = Unimpler	nented bit, read	l as '0'					
u = Bit is uncl	nanged	x = Bit is unki	nown	-n/n = Value a	at POR and BO	R/Value at all o	ther Resets				
'1' = Bit is set	-	'0' = Bit is cle	ared								
	000 (T. 0.)										
bit 7		e 3 Data 4 True	(non-inverted)) DIT							
	1 = d4T is g 0 = d4T is n	ated into g3 ot gated into g3	1								
bit 6		e 3 Data 4 Neg		bit							
	1 = d4N is g	•									
	-	d = d4N is not gated into g3									
bit 5	G3D3T: Gate	ate 3 Data 3 True (non-inverted) bit									
		1 = d3T is gated into g3									
		ot gated into g3									
bit 4		e 3 Data 3 Neg	ated (inverted)) bit							
	1 = d3N is g	pated into g3 not gated into g3	2								
bit 3		v) bit							
bit 5		Sate 3 Data 2 True (non-inverted) bit s gated into g3									
		ot gated into g3	6								
bit 2	G3D2N: Gat	e 3 Data 2 Neg	ated (inverted)) bit							
	1 = d2N is g	1 = d2N is gated into g3									
	0 = d2N is n	not gated into ga	3								
bit 1	G3D1T: Gate	G3D1T: Gate 3 Data 1 True (non-inverted) bit									
	1 = d1T is g 0 = d1T is n	ated into g3 ot gated into g3	6								
bit 0	G3D1N: Gat	e 3 Data 1 Neg	ated (inverted)) bit							
	1 = d1N is g										
	0 = d1N is n	not gated into g3	3								

REGISTER 28-9: CLCxGLS2: GATE 3 LOGIC SELECT REGISTER

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u				
G4D4T	G4D4N	G4D3T	G4D3N	G4D2T	G4D2N	G4D1T	G4D1N				
bit 7		•					bit				
Legend:											
R = Readable		W = Writable	bit	•	nented bit, read						
u = Bit is unc	hanged	x = Bit is unk	nown	-n/n = Value a	at POR and BO	R/Value at all o	ther Resets				
'1' = Bit is set		'0' = Bit is cle	ared								
bit 7	G4D4T: Gat	e 4 Data 4 True	(non-inverted)) bit							
	1 = d4T is g 0 = d4T is n	ated into g4 ot gated into g4	ŀ								
bit 6	G4D4N: Gat	te 4 Data 4 Neg	ated (inverted)) bit							
	1 = d4N is gated into g4										
		not gated into ga									
bit 5		G4D3T: Gate 4 Data 3 True (non-inverted) bit									
	1 = d3T is gated into g4 0 = d3T is not gated into g4										
bit 4		•) bit							
		G4D3N: Gate 4 Data 3 Negated (inverted) bit 1 = d3N is gated into g4									
	0 = d3N is not gated into g4										
bit 3	G4D2T: Gat	G4D2T: Gate 4 Data 2 True (non-inverted) bit									
	1 = d2T is g	1 = d2T is gated into $q4$									
	0 = d2T is n	ot gated into g4	Ļ								
bit 2	G4D2N: Gat	te 4 Data 2 Neg	ated (inverted)) bit							
		1 = d2N is gated into g4									
		0 = d2N is not gated into g4									
bit 1		G4D1T: Gate 4 Data 1 True (non-inverted) bit									
	1 = d1T is g	ated into g4 ot gated into g4									
		•) hit							
bit 0		te 4 Data 1 Neg									
	$1 = d \ln \log q$ $0 = d \ln \log q$	gated into g4									

REGISTER 28-10: CLCxGLS3: GATE 4 LOGIC SELECT REGISTER

REGISTER 28-11: CLCDATA: CLC DATA OUTPUT

U-0	U-0	U-0	U-0	R-0	R-0	R-0	R-0
—	—	—	—	MLC4OUT	MLC3OUT	MLC2OUT	MLC1OUT
bit 7							bit 0
Legend:							
R = Readable	R = Readable bit W = Writable bit		bit	U = Unimpler	nented bit, read	as '0'	
u = Bit is uncha	anged	x = Bit is unkr	iown	-n/n = Value at POR and BOR/Value		R/Value at all o	ther Resets
'1' = Bit is set		'0' = Bit is clea	ared				
bit 7-4	Unimplement	ed: Read as '0'					
bit 3	MLC4OUT: M	irror copy of LC	C4OUT bit				
bit 2	MLC3OUT: Mirror copy of LC3OUT bit						
bit 1	MLC2OUT: Mirror copy of LC2OUT bit						
bit 0	MLC1OUT: Mirror copy of LC1OUT bit						

Name	Bit7	Bit6	Bit5	Bit4	Blt3	Bit2	Bit1	Bit0	Register on Page
ANSELA	_	_	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0	177
ANSELB	_	_	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	182
ANSELC	ANSC7	ANSC6	ANSC5	ANSC4	ANSC3	ANSC2	—	—	187
CLCxCON	EN	_	OUT	INTP	INTN		MODE<2:0>		397
CLCDATA	_	_	_	_	MLC4OUT	MLC3OUT	MLC2OUT	MLC1OUT	404
CLCxGLS0	G1D4T	G1D4N	G1D3T	G1D3N	G1D2T	G1D2N	G1D1T	G1D1N	400
CLCxGLS1	G2D4T	G2D4N	G2D3T	G2D3N	G2D2T	G2D2N	G2D1T	G2D1N	401
CLCxGLS2	G3D4T	G3D4N	G3D3T	G3D3N	G3D2T	G3D2N	G3D1T	G3D1N	402
CLCxGLS3	G4D4T	G4D4N	G4D3T	G4D3N	G4D2T	G4D2N	G4D1T	G4D1N	403
CLCxPOL	POL	_	_	_	G4POL	G3POL	G2POL	G1POL	398
CLCxSEL0	_	—			D1S•	<5:0>			399
CLCxSEL1	—	—			D2S•	<5:0>			399
CLCxSEL2	_	—			D3S•	<5:0>			399
CLCxSEL3					D4S•	<5:0>			400
CLCINxPPS	-	—			CLCINxF	PS<5:0>			205, 207
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	132
PIE3	_	—	COG2IE	ZCDIE	CLC4IE	CLC3IE	CLC2IE	CLC1IE	135
PIR3	_	_	COG2IF	ZCDIF	CLC4IF	CLC3IF	CLC2IF	CLC1IF	141
RxyPPS	_	_			RxyPP	S<5:0>			205
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	176
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	181
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	186

TABLE 28-4:	SUMMARY OF REGISTERS ASSOCIATED WITH CLCx
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Legend: — = unimplemented read as '0'. Shaded cells are not used for CLC module.

29.0 OPERATIONAL AMPLIFIER (OPA) MODULES

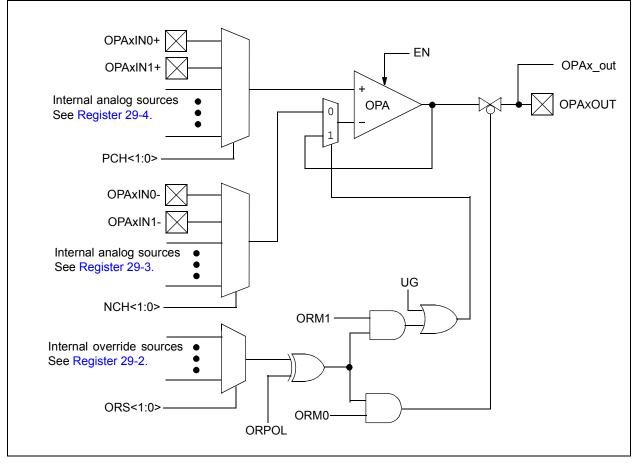
The Operational Amplifier (OPA) is a standard threeterminal device requiring external feedback to operate. The OPA module has the following features:

- External connections to I/O ports
- · Low leakage inputs
- Factory Calibrated Input Offset Voltage
- Unity gain control
- Programmable positive and negative source selections
- Override controls
 - Forced tri-state output
 - Forced unity gain

FIGURE 29-1: OPAx MODULE BLOCK DIAGRAM

TABLE 29-1: AVAILABLE OP AMP MODULES

Device	OPA1	OP2	OPA3	OPA4
PIC16(L)F1778	•	•	•	
PIC16(L)F1777/9	•	•	•	•



29.1 OPA Module Performance

Common AC and DC performance specifications for the OPA module:

- Common-Mode Voltage Range
- Leakage Current
- Input Offset Voltage
- Open-Loop Gain
- · Gain Bandwidth Product

Common-mode voltage range is the specified voltage range for the OPA+ and OPA- inputs, for which the OPA module will perform to within its specifications. The OPA module is designed to operate with input voltages between Vss and VDD. Behavior for commonmode voltages greater than VDD, or below Vss, are not guaranteed.

Leakage current is a measure of the small source or sink currents on the OPA+ and OPA- inputs. To minimize the effect of leakage currents, the effective impedances connected to the OPA+ and OPA- inputs should be kept as small as possible and equal.

Input offset voltage is a measure of the voltage difference between the OPA+ and OPA- inputs in a closed loop circuit with the OPA in its linear region. The offset voltage will appear as a DC offset in the output equal to the input offset voltage, multiplied by the gain of the circuit. The input offset voltage is also affected by the common-mode voltage. The OPA is factory calibrated to minimize the input offset voltage of the module.

Open-loop gain is the ratio of the output voltage to the differential input voltage, (OPA+) - (OPA-). The gain is greatest at DC and falls off with frequency.

Gain Bandwidth Product or GBWP is the frequency at which the open-loop gain falls off to 0 dB.

29.2 OPA Module Control

The OPA module is enabled by setting the OPAxEN bit of the OPAxCON register (Register 29-1). When enabled, the OPA forces the output driver of the OPAxOUT pin into tri-state to prevent contention between the driver and the OPA output.

Note: When the OPA module is enabled, the OPAxOUT pin is driven by the op amp output, not by the PORT digital driver. Refer to Table 36-17: Operational Amplifier (OPA) for the op amp output drive capability.

29.2.1 UNITY GAIN MODE

The OPAxUG bit of the OPAxCON register (Register 29-1) selects the Unity Gain mode. When unity gain is selected, the OPA output is connected to the inverting input and the OPAxIN pin is relinquished, releasing the pin for general purpose input and output.

29.2.2 PROGRAMMABLE SOURCE SELECTIONS

The inverting and non-inverting sources are selected with the OPAxNCHS (Register 29-3) and OPAxPCHS (Register 29-4) registers, respectively. Sources include:

- Internal DACs
- Device pins
- Internal slope compensation ramp generator
- Other op amps in the device

29.3 Override Control

29.3.1 OVERRIDE MODE

The op amp operation can be overridden in two ways:

- · Forced tri-state output
- · Force unity gain

The Override mode is selected with the ORM<1:0> bits of the OPxCON register (Register 29-1). The override is in effect when the mode is selected and the override source is true.

29.3.2 OVERRIDE SOURCES

The override source is selected with the OPAxORS register (Register 29-2). Sources are from internal peripherals including:

- CCP outputs
- · PWM outputs
- Comparator outputs
- · Zero-cross detect output
- Configurable Logic Cell outputs
- · COG outputs

29.3.3 OVERRIDE SOURCE POLARITY

The override source polarity can be inverted so that the override will occur on either the high or low level of the selected source. Override polarity is controlled by the ORPOL bit of the OPAxCON register (Register 29-1).

29.4 Effects of Reset

A device Reset forces all registers to their Reset state. This disables the OPA module.

29.5 Effects of Sleep

The operational amplifier continues to operate when the device is put in Sleep mode.

29.6 Register Definitions: Op Amp Control

Long bit name prefixes for the op amp peripherals are shown in Table 29-2. Refer to **Section 1.1 "Register and Bit naming conventions**" for more information

TABLE 29-2:

Peripheral	Bit Name Prefix
OPA1	OPA1
OPA2	OPA2
OPA3	OPA3
OPA4 ⁽¹⁾	OPA4

Note 1: PIC16(L)F1777/9 only.

REGISTER 29-1: OPAxCON: OPERATIONAL AMPLIFIER (OPAx) CONTROL REGISTER

R/W-0/0	U-0	U-0	R/W-0/0	U-0	R/W-0/0	R/W-0/0	R/W-0/0
EN	—	—	UG	—	ORPOL	ORM	<1:0>
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 7	EN: Op Amp Enable bit
	1 = Op amp is enabled
	0 = Op amp is disabled and consumes no active power
bit 6-5	Unimplemented: Read as '0'
bit 4	UG: Op Amp Unity Gain Select bit
	 1 = OPA output is connected to inverting input. OPAxIN- pin is available for general purpose I/O. 0 = Inverting input is connected to the OPAxIN- pin
bit 3	Unimplemented: Read as '0'
bit 2	ORPOL: Op Amp Override Source Polarity bit
	1 = Override source polarity is inverted. Override occurs when source is high.
	0 = Override source polarity is not inverted. Override occurs when source is low.
bit 1-0	ORM<1:0>: Op Amp Override Mode Selection bits
	11 = Reserved. Do not use.
	10 = Op amp is forced to unity gain when override source is true.
	01 = Op amp output is tri-stated when override source is true.
	00 = Output override function is disabled.

REGISTER 29-2: OPAXORS: OP AMP OVERRIDE SOURCE SELECTION REGISTER

U-0	U-0	U-0	R/W-0/0	R/W-0/x	R/W-0/x	R/W-0/0	R/W-0/x
—	_	-			ORS<4:0>		
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 7-5 Unimplemented: Read as '0'

bit 4-0 ORS<4:0>: Op Amp Output Override Source Selection bits See Table 29-3: Override Sources

TABLE 29-3: OVERRIDE SOURCES

ORS<4:0>	OPA1	OPA2	OPA3	OPA4 ⁽¹⁾
11111	COG2D	COG2D	COG4D ⁽¹⁾	COG4D
11110	COG2C	COG2C	COG4C ⁽¹⁾	COG4C
11101	COG2B	COG2B	COG4B ⁽¹⁾	COG4B
11100	COG2A	COG2A	COG4A ⁽¹⁾	COG4A
11011	COG1D	COG1D	COG3D	COG3D
11010	COG1C	COG1C	COG3C	COG3C
11001	COG1B	COG1B	COG3B	COG3B
11000	COG1A	COG1A	COG3A	COG3A
10111	LC4_out	LC4_out	LC4_out	LC4_out
10110	LC3_out	LC3_out	LC3_out	LC3_out
10101	LC2_out	LC2_out	LC2_out	LC2_out
10100	LC1_out	LC1_out	LC1_out	LC1_out
10011	sync_C8OUT ⁽¹⁾	sync_C8OUT ⁽¹⁾	sync_C8OUT ⁽¹⁾	sync_C8OUT
10010	sync_C7OUT ⁽¹⁾	sync_C7OUT ⁽¹⁾	sync_C7OUT ⁽¹⁾	sync_C7OUT
10001	sync_C6OUT	sync_C6OUT	sync_C6OUT	sync_C6OUT
10000	sync_C5OUT	sync_C5OUT	sync_C5OUT	sync_C5OUT
01111	sync_C4OUT	sync_C4OUT	sync_C4OUT	sync_C4OUT
01110	sync_C3OUT	sync_C3OUT	sync_C3OUT	sync_C3OUT
01101	sync_C2OUT	sync_C2OUT	sync_C2OUT	sync_C2OUT
01100	sync_C1OUT	sync_C1OUT	sync_C1OUT	sync_C1OUT
01011	PWM12_out ⁽¹⁾	PWM12_out ⁽¹⁾	PWM12_out ⁽¹⁾	PWM12_out
01010	PWM11_out	PWM11_out	PWM11_out	PWM11_out
01001	PWM6_out	PWM6_out	PWM6_out	PWM6_out
01000	PWM5_out	PWM5_out	PWM5_out	PWM5_out
00111	PWM10_out ⁽¹⁾	PWM10_out ⁽¹⁾	PWM10_out ⁽¹⁾	PWM10_out
00110	PWM9_out	PWM9_out	PWM9_out	PWM9_out
00101	PWM4_out	PWM4_out	PWM4_out	PWM4_out
00100	PWM3_out	PWM3_out	PWM3_out	PWM3_out
00011	CCP8_out ⁽¹⁾	CCP8_out ⁽¹⁾	CCP8_out ⁽¹⁾	CCP8_out
00010	CCP7_out	CCP7_out	CCP7_out	CCP7_out
00001	CCP2_out	CCP2_out	CCP2_out	CCP2_out
00000	CCP1_out	CCP1_out	CCP1_out	CCP1_out

Note 1: PIC16(L)F1777/9 only.

U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	—	—	-		NCH	<3:0>	
bit 7				- -			bit 0
Legend:							

=ogonan		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 7-4 Unimplemented: Read as '0'

bit 3-0 NCH<3:0>: Op Amp Inverting Input Channel Selection bits See Table 29-4: Inverting Input Sources

TABLE 29-4: INVERTING INPUT SOURCES

NCH<3:0>	OPA1	OPA2	OPA3	OPA4 ⁽¹⁾
1111	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use
1110	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use
1101	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use
1100	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use
1011	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use
1010	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use
1001	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use
1000	PRG2_out	PRG2_out	PRG4_out ⁽¹⁾	PRG4_out
0111	PRG1_out	PRG1_out	PRG3_out	PRG3_out
0110	FVR_Buffer1	FVR_Buffer1	FVR_Buffer2	FVR_Buffer2
0101	DAC4_out	DAC4_out	DAC8_out ⁽¹⁾	DAC8_out
0100	DAC3_out	DAC3_out	DAC7_out	DAC7_out
0011	DAC2_out	DAC2_out	DAC6_out ⁽¹⁾	DAC6_out
0010	DAC1_out	DAC1_out	DAC5_out	DAC5_out
0001	OPA1IN1-	OPA2IN1-	OPA3IN1- ⁽¹⁾	OPA4IN1-
0000	OPA1IN0-	OPA2IN0-	OPA3IN0-	OPA3IN0-

Note 1: PIC16(L)F1777/9 only.

REGISTER 29-4: OPAxPCHS: OP AMP POSITIVE CHANNEL SOURCE SELECT REGISTER

U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	—		—		PCH	<3:0>	
bit 7							bit 0
Legend:							

0		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 7-4 Unimplemented: Read as '0'

bit 3-0 PCH<3:0>: Op Amp Non-Inverting Input Channel Selection bits See Table 29-5: Non-Inverting Input Sources

TABLE 29-5: NON-INVERTING INPUT SOURCES

NCH<3:0>	OPA1	OPA2	OPA3	OPA4 ⁽¹⁾
1111	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use
1110	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use
1101	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use
1100	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use
1011	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use
1010	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use
1001	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use	Reserved. Do not use
1000	PRG2_out	PRG2_out	PRG4_out ⁽¹⁾	PRG4_out
0111	PRG1_out	PRG1_out	PRG3_out	PRG3_out
0110	FVR_Buffer1	FVR_Buffer1	FVR_Buffer2	FVR_Buffer2
0101	DAC4_out	DAC4_out	DAC8_out ⁽¹⁾	DAC8_out
0100	DAC3_out	DAC3_out	DAC7_out	DAC7_out
0011	DAC2_out	DAC2_out	DAC6_out ⁽¹⁾	DAC6_out
0010	DAC1_out	DAC1_out	DAC5_out	DAC5_out
0001	OPA1IN1+	OPA2IN1+	OPA3IN1+ ⁽¹⁾	OPA4IN1+
0000	OPA1IN0+	OPA2IN0+	OPA3IN0+	OPA4IN0+

Note 1: PIC16(L)F1777/9 only.

PIC16(L)F1777/8/9

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ANSELB			ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	182
ANSELC	ANSC7	ANSC6	ANSC5	ANSC4	ANSC3	ANSC2	_	_	187
DAC1CON0	EN	FM	OE1	OE2	PSS		NSS	ـــــــــــــــــــــــــــــــــــــ	249
DAC2CON0	EN	FM	OE1	OE2	PSS<	<1:0>	NSS	<1:0>	249
DAC5CON0	EN	FM	OE1	OE2	PSS<	<1:0>	NSS	<1:0>	249
DAC3CON0	EN	_	OE1	OE2	PSS<	<1:0>	NSS	<1:0>	244
DAC4CON0	EN	_	OE1	OE2	PSS<	<1:0>	NSS	<1:0>	244
DAC7CON0	EN	_	OE1	OE2	PSS<	<1:0>	NSS	<1:0>	244
DAC3REF				REF<4:0>					
DAC4REF						REF<4:0>			245
DAC7REF						REF<4:0>			245
DAC1REFH			R	EF<9:x> (x D	epends on FM	bit)			250
DAC2REFH			R	EF<9:x> (x D	epends on FM	bit)			250
DAC5REFH			R	EF<9:x> (x D	epends on FM	bit)			250
FVRCON	FVREN	FVRRDY	TSEN	TSRNG	CDAFV	′R<1:0>	ADFVI	R<1:0>	223
OPAxCON	EN	_	_	UG	—	ORPOL	ORM	<1:0>	408
OPAxNCHS	_	_	_	_		NCH<	3:0>		410
OPAxPCHS	_	_	_	_		PCH<	3:0>		411
OPAxORS	_					ORS<4:0>			409
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	181
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	186

TABLE 29-6: SUMMARY OF REGISTERS ASSOCIATED WITH OP AMPS

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by op amps.

30.0 PROGRAMMABLE RAMP GENERATOR (PRG) MODULE

The Programmable Ramp Generator (PRG) module is designed to provide rising and falling linear ramps. Typical applications include slope compensation for fixed frequency, continuous current, and Current mode switched power supplies. Slope compensation is a necessary feature of these power supplies because it prevents frequency instabilities at duty cycles greater than 50%.

The PRG has the following features:

- · Linear positive and negative voltage ramp outputs
- Programmable current source/sink
- Internal and external reference voltage selection
- Internal and external timing source selection

A simplified block diagram of the PRG is shown in Figure 30-1.

Device	PRG1	PRG2	PRG3	PRG4
PIC16(L)F1778	•	•	•	
PIC16(L)F1777/9	٠	٠	•	•

TABLE 30-1: AVAILABLE PRG MODULES

30.1 Fundamental Operation

The PRG can be operated in three voltage ramp generator modes:

- Falling Voltage (slope compensation)
- Rising Voltage
- Alternating Rising and Falling Voltage

In the Rising or Falling mode an internal capacitor is discharged when the set_falling timing input is true and charged by an internally generated constant current when the set_rising timing input is true. The resulting linear ramp starts at the selected voltage input level and resets back to that level when the ramp is terminated by the set_falling timing input. The set_falling input dominates when both timing inputs are true.

To control the operation with a single-ended source, select the same source for both the set_rising and set_falling inputs and invert the polarity of one of them with the corresponding polarity control bit.

In the Alternating mode the capacitor is not discharged but alternates between being charged in one direction then the other.

Input selections are identical for all modes. The input voltage is supplied by any of the following:

- The PRGxIN0 or PRGxIN1 pins
- The buffered output of the internal Fixed Voltage Reference (FVR),
- Any of the internal DACs.

The timing sources are selected from the following:

- · The synchronized output of any comparator
- Any PWM output
- Any I/O pin

The ramp output is available as an input to any of the comparators or op amps.

30.1.1 SLOPE COMPENSATION

Slope compensation works by quickly discharging an internal capacitor at the beginning of each PWM period. One side of the internal capacitor is connected to the voltage input source and the other side is connected to the internal current sink. The internal current sink charges this capacitor at a programmable rate. As the capacitor charges, the capacitor voltage is subtracted from the voltage source, producing a linear voltage decay at the required rate (see Figure 30-2). The ramp terminates and the capacitor is discharged when the set_falling timing input goes true.

Enabling the optional one-shot by setting the OS bit of the PRGxCON0 register ensures that the capacitor is fully discharged by overriding the set_rising timing input and holding the shorting switch closed for at least the one-shot period, typically 50 ns. Edge sensitive timing inputs that occur during the one-shot period will be ignored. Level sensitive timing inputs that occur during, and extend beyond, the one-shot period will be suspended until the end of the one-shot time.

30.1.2 RAMP GENERATION

Ramp generation is similar to slope compensation except that the slope is either both rising and falling or just rising.

30.1.2.1 Alternating Rising/Falling Ramps

The alternating rising/falling ramp generation function works by employing the built-in current source and sink and relying on the synchronous control of the internal analog switches and timing sources to ramp the module's output voltage up and then subsequently down.

Once initialized, the output voltage is ramped up linearly by the current source at a programmable rate until the set_falling timing source goes true, at which point the current source is disengaged. At the same time, the current sink is engaged to linearly ramp down the output voltage, also at a programmable rate, until the set_rising timing input goes true, thereby reversing the ramp slope. The process then repeats to create a saw tooth like waveform as shown in Figure 30-3 and Figure 30-4.

The set_rising and set_falling timing inputs can be either edge or level sensitive which is selected with the respective REDG and FEDG bits of the PRGxCON0 register. Edge sensitive operation is recommended for periodic signals such as clocks, and level sensitive operation is recommended for analog limit triggers such as comparator outputs.

When the one-shot is enabled (OS bit is set) then both the falling and rising ramps will persist for a minimum of the one-shot period. Edge sensitive timing inputs that occur during the one-shot period will be ignored. Level sensitive timing inputs that occur during, and extend beyond, the one-shot period will be suspended until the end of the one-shot time.

30.1.2.2 Rising Ramp

The Rising Ramp mode is identical to the Slope Compensation mode except that the ramps have a rising slope instead of a falling slope. One side of the internal capacitor is connected to the voltage input source and the other side is connected to the internal current source. The internal current source charges this capacitor at a programmable rate. As the capacitor charges, the capacitor voltage is added to the voltage source, producing a linear voltage rise at the required rate (see Figure 30-5). The ramp terminates and the capacitor is discharged when the set_falling timing input goes true. The next ramp starts when the set_rising timing input goes true.

Enabling the optional one-shot by setting the OS bit of the PRGxCON0 register ensures that the capacitor is fully discharged by overriding the set_rising timing input and holding the shorting switch closed for at least the one-shot period, typically 50 ns. Edge sensitive timing inputs that occur during the one-shot period will be ignored. Level sensitive timing inputs that occur during, and extend beyond, the one-shot period will be suspended until the end of the one-shot time.

30.2 Enable, Ready, Go

The EN bit of the PRGxCON0 register enables the analog circuitry including the current sources. This permits preparing the PRG module for use and allowing it to become stable before putting it into operation. When the EN bit is set then the timing inputs are enabled so that initial ramp action can be determined before the GO bit is set. The capacitor shorting switch is closed when the EN bit is set and remains closed while the GO bit is zero.

The RDY bit of the PRGxCON1 register indicates that the analog circuits and current sources are stable.

The GO bit of the PRGxCON0 register enables the switch control circuits, thereby putting the PRG into operation. The GO transition from cleared to set triggers the one-shot, thereby extending the capacitor shorting switch closure for the one-shot period.

To ensure predictable operation, set the EN bit first then wait for the RDY bit to go high before setting the GO bit.

30.3 Independent Set_rising and Set_falling Timing Inputs

The timing inputs determine when the ramp starts and stops. In the Alternating Rising/Falling mode the ramp rises when the set_rising input goes true and falls when the set_falling input goes true. In the Slope Compensation and Rising Ramp modes the capacitor is discharged when the set_falling timing input goes true and the ramp starts when the set_rising timing input goes true. The set_falling input dominates the set_rising input.

30.4 Level and Edge Timing Sensitivity

The set_rising and set_falling timing inputs can be independently configured as either level or edge sensitive.

Level sensitive operation is useful when it is necessary to detect a timing input true state after an overriding condition ceases. For example, level sensitivity is useful for capacitor generated timing inputs that may be suppressed by the overriding action of the one-shot. With level sensitivity a capacitor output that changes during the one-shot period will be detected at the end of the one-shot time. With edge sensitivity the change would be ignored.

Edge sensitive operation is useful for periodic timing inputs such as those generated by PWMs and clocks. The duty cycle of a level sensitive periodic signal may interfere with the other timing input. Consider an Alternating Ramp mode with a level sensitive 50% PWM as the set_rising timing source and a level sensitive comparator as the set_falling timing source. If the comparator output reverses the ramp while the PWM signal is still high then the ramp will improperly reverse again when the comparator signal goes low. That same scenario with the set_rising timing input set for edge sensitivity would properly change the ramp output to rising only on the rising edge of the PWM signal.

Set_rising and set_falling timing input edge sensitivity is selected with the respective REDG and FEDG bits of the PRGxCON1 register.

30.5 One-Shot Minimum Timing

The one-shot timer ensures a minimum capacitor discharge time in the Slope Compensation and Rising Ramp modes, and a minimum rising or falling ramp duration in the Alternating Ramp mode. Setting the OS bit of the PRGxCON0 register enables the one-shot timer.

30.6 DAC Voltage Sources

When using any of the DACs as the voltage source expect a voltage offset equal to the current setting times the DAC equivalent resistance. This will be a constant offset in the Slope Compensation and Ramp modes and a positive/negative step offset in the Alternating mode. To avoid this limitation, feed the DAC output to the PRG input through one of the op amps set for unity gain.

30.7 Operation During Sleep

The RG module is unaffected by Sleep.

30.8 Effects of a Reset

The RG module resets to a disabled condition.

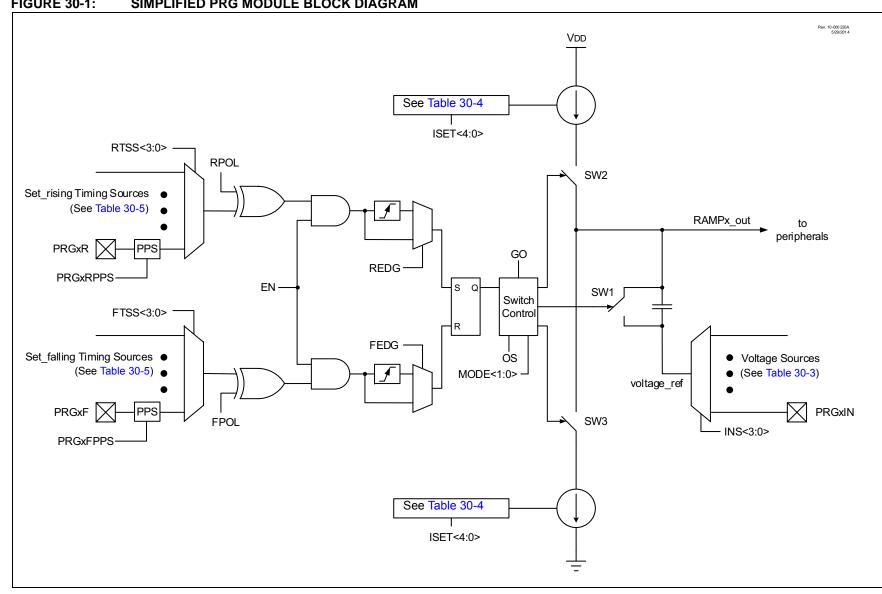


FIGURE 30-1:

SIMPLIFIED PRG MODULE BLOCK DIAGRAM

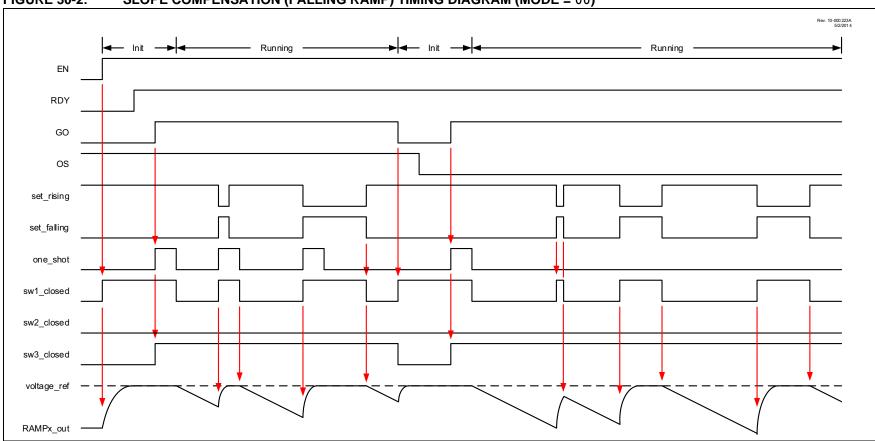


FIGURE 30-2:

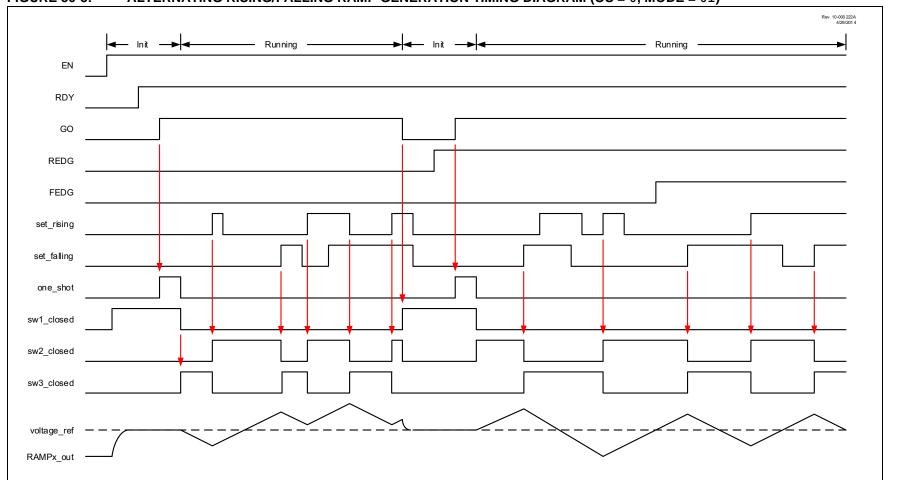
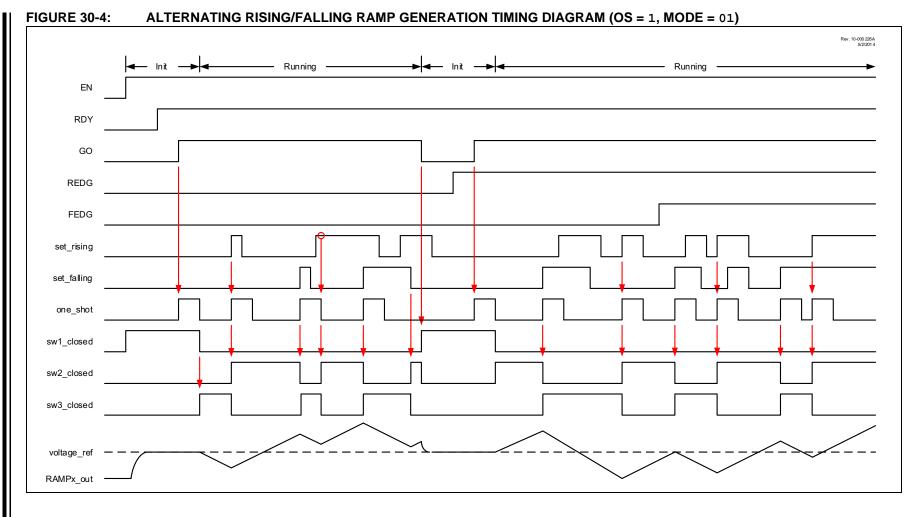
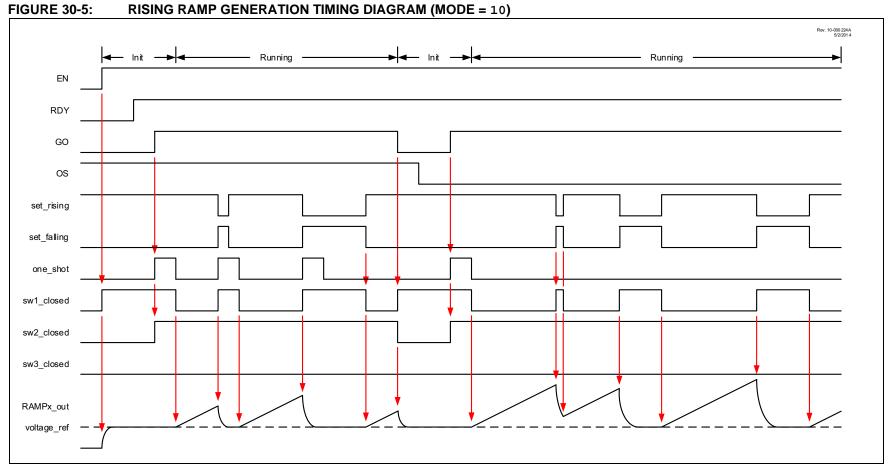


FIGURE 30-3: ALTERNATING RISING/FALLING RAMP GENERATION TIMING DIAGRAM (OS = 0, MODE = 01)



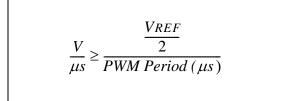


RISING RAMP GENERATION TIMING DIAGRAM (MODE = 10)

30.9 Slope Compensation Application

An example slope compensation circuit is shown in Figure 30-6. The PRG input voltage is PRGxIN which shares an I/O pin with the op amp output. The op amp output is designed to operate at the expected peak current sense voltage (i.e., VREF). The PRG output voltage starts at VREF and should fall at a rate less than half the target circuit current sense voltage rate of rise. Therefore, the compensator slope expressed as volts per µs can be computed by Equation 30-1.

EQUATION 30-1:



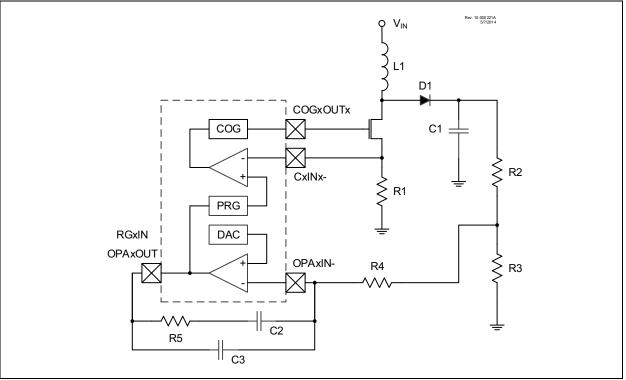
For example, when the circuit is using a 1Ω current sense resistor and the peak current is 1A, then the peak current expressed as a voltage is 1V. Therefore, for this example, the op amp output should be designed to operate at 1V. If the power supply PWM frequency is 1 MHz, then the period is 1 μ s. Therefore, the desired slope is 0.5 V/ μ s, which is computed as shown in Equation 30-2.

EQUATION 30-2:

$$\frac{\frac{VREF}{2}}{PWM Period (\mu s)} = \frac{\frac{1}{2}}{1\mu s} = 0.5 V/\mu s$$

Note: The setting for $0.5V/\mu s$ is ISET<4:0> = 6

FIGURE 30-6: EXAMPLE SLOPE COMPENSATION CIRCUIT



30.10 Register Definitions: Slope Compensation Control

Long bit name prefixes for the PRG peripherals are shown in Table 30-2. Refer to **Section 1.1 "Register and Bit naming conventions**" for more information **TABLE 30-2:**

Peripheral	Bit Name Prefix		
PRG1	RG1		
PRG2	RG2		
PRG3	RG3		
PRG4 ⁽¹⁾	RG4		

Note 1: PIC16(L)F1777/9 only.

REGISTER 30-1: PRGxCON0: PROGRAMMABLE RAMP GENERATOR CONTROL 0 REGISTER

R/W-0/0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
EN	—	FEDG	REDG	MOD	E<1:0>	OS	GO
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = value depends on configuration bits

bit 7	 EN: Programmable Ramp Generator Enable bit 1 = PRG module is enabled 0 = PRG module is disabled
bit 6	Unimplemented: Read as '0'
bit 5	FEDG: Set_falling Input Mode Select bit
	1 = Set_falling timing input is edge sensitive0 = Set_falling timing input is level sensitive
bit 4	REDG: Set_rising Input Mode Select bit
	1 = Set_rising timing input is edge sensitive0 = Set_rising timing input is level sensitive
bit 3-2	<pre>MODE<1:0>: Programmable Ramp Generator Mode Selection bits 11 = Reserved 10 = Rising Ramp Generator 01 = Alternating Rising/Falling Ramp Generator 00 = Slope Compensation</pre>
bit 1	OS: One-Shot Enable bit
	 1 = One-shot is enabled. Minimum capacitor discharge is internally timed by one-shot. 0 = One-shot is disabled. Capacitor is discharged when timing input is true.
bit O	 GO: Ramp Generation Control Start bit <u>if EN = 1</u>: 1 = Slope or Ramp function is operating 0 = Slope or Ramp function is not operating. All current source current source switches are open and capacitor discharge switch is closed. <u>If EN = 0</u>: This bit is forced to 0

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REGISTER 30-2: PRGxCON1: PROGRAMMABLE RAMP GENERATOR CONTROL 1 REGISTER

U-0	U-0	U-0	U-0	U-0	R-0	R/W-0/0	R/W-0/0	
	—	_	_	—	RDY	FPOL	RPOL	
bit 7							bit	
Legend:								
R = Readable	bit	W = Writable b	it	U = Unimpleme	ented bit, read a	s 'O'		
u = Bit is unch	anged	x = Bit is unknown		-n/n = Value at	POR and BOR/	Value at all other I	Resets	
'1' = Bit is set		'0' = Bit is clear	red	q = value depends on configuration bits				
bit 7-3	Unimplemente	d: Read as '0'						
bit 2	1 = PRG is rea	RDY: Slope Generator Ready Status bit 1 = PRG is ready 0 = PRG is not ready						
bit 1	FPOL: Fall Event Polarity Select bit 1 = Set_falling timing input is active-low 0 = Set_falling timing input is active-high							
bit 0	RPOL: Rise Ev	ent Polarity Sele	ect bit					

REGISTER 30-3: PRGxINS: VOLTAGE INPUT SELECT REGISTER

1 = Set_rising timing input is active-low0 = Set_rising timing input is active-high

U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	
—	_	—	—	INS<3:0>				
bit 7 bit 0								

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = value depends on configuration bits

bit 7-4 Unimplemented: Read as '0'

bit 3-0 **INS<3:0>:** Voltage Input Select bits Selects source of voltage level at which the ramp starts. See Table 30-3.

TABLE 30-3: VOLTAGE INPUT SOURCES

INS<2:0>	PRG1 Voltage Source	PRG2 Voltage Source	PRG3 Voltage Source	PRG4 Voltage Source ⁽²⁾			
1010-1111	Reserved	Reserved	Reserved	Reserved			
1001 (1)	Switched PRG1IN1/OPA2OUT	Switched PRG1IN1/OPA2OUT	Switched PRG3IN1/OPA4OUT ⁽²⁾	Switched PRG4IN1/OPA3OUT			
1000 (1)	Switched PRG1IN0/OPA1OUT	Switched PRG1IN0/OPA1OUT	Switched PRG3IN0/OPA3OUT	Switched PRG4IN0/OPA4OUT			
0111	Reserved	Reserved	Reserved	Reserved			
0110	DAC4_output	DAC4_output	DAC8_output ⁽²⁾	DAC8_output			
0101	DAC3_output	DAC3_output	DAC7_output	DAC7_output			
0100	DAC2_output	DAC2_output	DAC6_output ⁽²⁾	DAC6_output			
0011	DAC1_output	DAC1_output	DAC5_output	DAC5_output			
0010	FVR_buffer1	FVR_buffer1	FVR_buffer2	FVR_buffer2			
0001	PRG1IN1/OPA2OUT	PRG2IN1/OPA1OUT	PRG3IN1/OPA4OUT ⁽²⁾	PRG4IN1/OPA3OUT			
0000	PRG1IN0/OPA1OUT	PRG2IN0/OPA2OUT	PRG3IN0/OPA3OUT	PRG4IN0/OPA4OUT			
Note 4. Insultaneous is suitabled off where an own substitute is farsing bid state. One Continue 00.0 (Opportunity							

Note 1: Input source is switched off when op amp override is forcing tri-state. See Section 29.3 "Override

Control". 2: PIC16(L)F1777/9 only.

-n/n = Value at POR and BOR/Value at all other Resets

q = value depends on configuration bits

U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	—	—			ISET<4:0>		
bit 7							bit 0
Legend:							
R = Readable bit W = Writable bit			bit	U = Unimpler	nented bit, read	as '0'	

REGISTER 30-4: PRGxCON2: PROGRAMMABLE RAMP GENERATOR CONTROL 2 REGISTER

bit 7-5 Unimplemented: Read as '0'

u = Bit is unchanged

'1' = Bit is set

bit 4-0 **ISET<4:0>:** PRG Current Source/Sink Set bits Current source/sink setting and slope rate. See Table 30-4.

x = Bit is unknown

'0' = Bit is cleared

TABLE 30-4: PROGRAMMABLE RAMP GENERATOR CURRENT SETTINGS

ISET<4:0>	Current Setting (uA)	Slope Rate (V/us)	ISET<4:0>	Current Setting (uA)	Slope Rate (V/us)
0h	2	0.2	10h	10	1.0
1h	2.5	0.25	11h	11	1.1
2h	3	0.3	12h	12	1.2
3h	3.5	0.35	13h	13	1.3
4h	4	0.4	14h	14	1.4
5h	4.5	0.45	15h	15	1.5
6h	5	0.5	16h	16	1.6
7h	5.5	0.55	17h	17	1.7
8h	6	0.6	18h	18	1.8
9h	6.5	0.65	19h	19	1.9
Ah	7	0.7	1Ah	20	2.0
Bh	7.5	0.75	1Bh	21	2.1
Ch	8	0.8	1Ch	22	2.2
Dh	8.5	0.85	1Dh	23	2.3
Eh	9	0.9	1Eh	24	2.4
Fh	9.5	0.95	1Fh	25	2.5

PIC16(L)F1777/8/9

U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	—	—	—		RTSS	<3:0>	
bit 7							bit 0
Legend:	Legend:						
R = Readable bit W = Writable bit		U = Unimplen	nented bit, read	as '0'			
u = Bit is unchanged x = Bit is unknown		-n/n = Value at POR and BOR/Value at all other Resets			ther Resets		
'1' = Bit is set '0' = Bit is cleared		q = value depends on configuration bits					

REGISTER 30-5: PRGxRTSS: SET_RISING TIMING SOURCE SELECT REGISTER

bit 7-4 Unimplemented: Read as '0'

bit 3-0 RTSS<3:0>: Set_rising Timing Source Select bits See Table 30-5.

REGISTER 30-6: PRGxFTSS: SET_FALLING TIMING SOURCE SELECT REGISTER

U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	—	—	_		FTSS	<3:0>	
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = value depends on configuration bits

bit 7-4 Unimplemented: Read as '0'

bit 3-0 FTSS<3:0>: Set_falling Timing Source Select bits See Table 30-5.

RTSS<3:0>/ FTSS<3:0>	PRG1 Timing Source	PRG2 Timing Source	PRG3 Timing Source	PRG4 Timing Source ⁽²⁾
1111	Reserved	Reserved	PWM12_output ⁽²⁾	PWM12_output
1110	Reserved	Reserved	PWM11_output	PWM11_output
1101	LC2_out	LC2_out	PWM6_output	PWM6_output
1100	LC1_out	LC1_out	PWM5_output	PWM5_output
1011	PWM10_output ⁽²⁾	PWM10_output ⁽²⁾	Reserved	Reserved
1010	PWM9_output	PWM9_output	Reserved	Reserved
1001	PWM4_output	PWM4_output	LC4_out ⁽²⁾	LC4_out
1000	PWM3_output	PWM3_output	LC3_out	LC3_out
0111	PRGxR/PRGxF Pin ⁽¹⁾	PRGxR/PRGxF Pin ⁽¹⁾	PRGxR/PRGxF Pin ⁽¹⁾	PRGxR/PRGxF Pin ⁽¹⁾
0110	Reserved	Reserved	sync_C7OUT ⁽²⁾	sync_C7OUT
0101	Reserved	Reserved	sync_C6OUT	sync_C6OUT
0100	Reserved	Reserved	sync_C5OUT	sync_C5OUT
0011	sync_C4OUT	sync_C4OUT	Reserved	Reserved
0010	sync_C3OUT	sync_C3OUT	Reserved	Reserved
0001	sync_C2OUT	sync_C2OUT	Reserved	Reserved
0000	sync_C1OUT	sync_C1OUT	Reserved	Reserved

TABLE 30-5 :	PROGRAMMABLE RAMP GENERATOR TIMING SOURCES
---------------------	--

Note 1: Input pin is selected with the PRGxRPPS or PRGxFPPS register.

2: PIC16(L)F1777/9 only.

PIC16(L)F1777/8/9

TABLE 30-6-	SUMMARY OF REGISTERS ASSOCIATED WITH THE PRG MODULE ⁽¹⁾
TADLE 30-0.	SOMMART OF REGISTERS ASSOCIATED WITH THE FRG MODULE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
PRGxCON0	EN	—	FEDG	REDG	MODE	=<1:0>	OS	GO	421
PRGxCON1	—	_	_	_	_	RDY	FPOL	RPOL	422
PRGxCON2	_	_	_			ISET<4:0>			423
PRGxINS	_	_	_	INS<3:0>					422
PRGxRPPS	_	_		PRGxRPPS<5:0>					424
PRGxFPPS	_	_		PRGxFPPS<5:0>				424	
PRGxRTSS	_	_	_	RTSS<3:0>					205, 207
PRGxFTSS	_	_	_	FTSS<3:0>				205, 207	
PORTC	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0	186
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	186
ANSELC	ANSC7	ANSC6	ANSC5	ANSC4	ANSC3	ANSC2	—	_	187
WPUC	WPUC7	WPUC6	WPUC5	WPUC4	WPUC3	WPUC2	WPUC1	WPUC0	188

Legend: — = unimplemented, read as '0'. Shaded cells are unused by the PRG module.

Note 1: PRG4 available on PIC16(L)F1777/9 only.

31.0 DATA SIGNAL MODULATOR (DSM)

The Data Signal Modulator (DSM) is a peripheral that allows the user to mix a data stream, also known as a modulator signal, with a carrier signal to produce a modulated output.

Both the carrier and the modulator signals are supplied to the DSM module either internally, from the output of a peripheral, or externally through an input pin.

The modulated output signal is generated by performing a logical "AND" operation of both the carrier and modulator signals and then provided to the MDxOUT pin.

The carrier signal is comprised of two distinct and separate signals. A carrier high (CARH) signal and a carrier low (CARL) signal. During the time in which the modulator (MOD) signal is in a logic high state, the DSM mixes the carrier high signal with the modulator signal. When the modulator signal is in a logic low state, the DSM mixes the carrier low signal with the modulator signal.

Using this method, the DSM can generate the following types of Key Modulation schemes:

- Frequency-Shift Keying (FSK)
- Phase-Shift Keying (PSK)
- On-Off Keying (OOK)

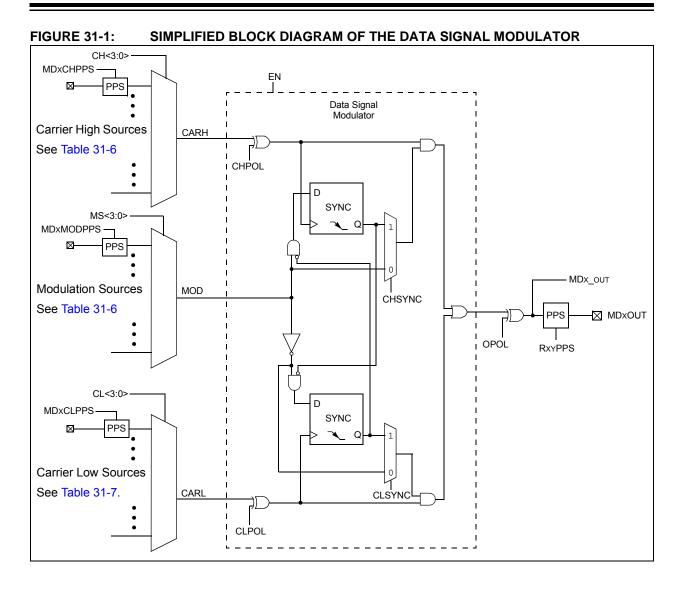
Additionally, the following features are provided within the DSM module:

- Carrier Synchronization
- Carrier Source Polarity Select
- Carrier Source Pin Disable
- Programmable Modulator Data
- Modulator Source Pin Disable
- · Modulated Output Polarity Select
- · Slew Rate Control

Figure 31-1 shows a Simplified Block Diagram of the Data Signal Modulator peripheral.

Device	DSM1	DSM2	DSM3	DSM4
PIC16(L)F1778	•	•	•	
PIC16(L)F1777/9	•	•	•	•

TABLE 31-1: AVAILABLE DSM MODULES



31.1 DSM Operation

The DSM module is enabled by setting the EN bit in the MDxCON register. Clearing the EN bit in the MDxCON register disables the DSM module by automatically switching the carrier high and carrier low signals to the Vss signal source. The modulator signal source is also switched to the BIT bit in the MDxCON0 register. This not only assures that the DSM module is inactive, but that it is also consuming the least amount of current.

The values used to select the carrier high, carrier low, and modulator sources held by the Modulation Source, Modulation High Carrier, and Modulation Low Carrier control registers are not affected when the EN bit is cleared and the DSM module is disabled. The values inside these registers remain unchanged while the DSM is inactive. The sources for the carrier high, carrier low and modulator signals will once again be selected when the EN bit is set and the DSM module is enabled and active.

The modulated output signal can be output on any device I/O pin by selecting the desired DSM module in the pin's PPS control register (see Register 12-2). If the output is not directed to any I/O pin then the DSM module will remain active and continue to mix signals, but the output value will not be sent to any pin.

31.2 Modulator Signal Sources

The modulator signal is selected by configuring the MS<4:0> bits of the MDxSRC register. Selections are shown in Table 31-6.

31.3 Carrier Signal Sources

The carrier high signal is selected by configuring the CH<4:0> bits of the MDxCARH register. Selections are shown in Table 31-6.

The carrier low signal is selected by configuring the CL<4:0> bits of the MDxCARL register. Selections are shown in Table 31-7.

31.4 Carrier Synchronization

During the time when the DSM switches between carrier high and carrier low signal sources, the carrier data in the modulated output signal can become truncated. To prevent this, the carrier signal can be synchronized to the modulator signal. When synchronization is enabled, the carrier pulse that is being mixed at the time of the transition is allowed to transition low before the DSM switches over to the next carrier source.

Synchronization is enabled separately for the carrier high and carrier low signal sources. Synchronization for the carrier high signal is enabled by setting the CHSYNC bit of the MDxCON1 register. Synchronization for the carrier low signal is enabled by setting the CLSYNC bit of the MDxCON1 register.

Figure 31-1 through Figure 31-6 show timing diagrams of using various synchronization methods.

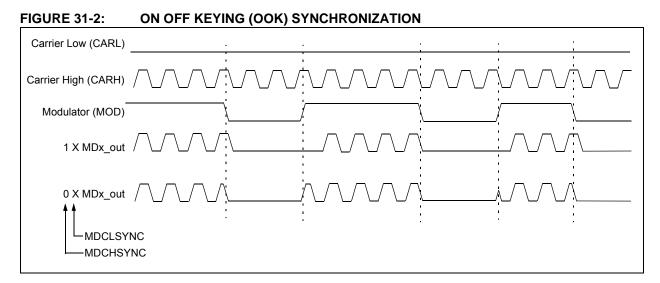
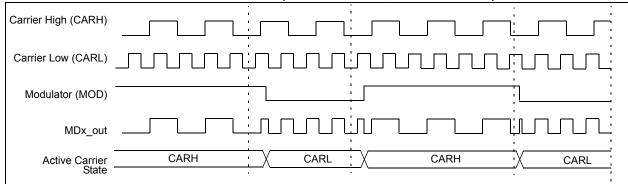


FIGURE 31-3: NO SYNCHRONIZATION (MDCHSYNC = 0, MDCLSYNC = 0)



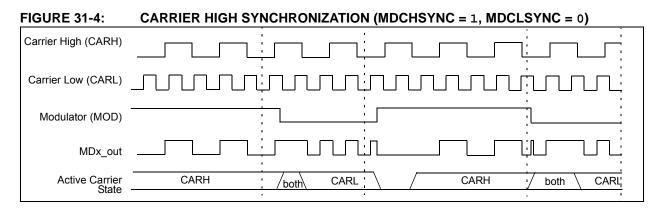
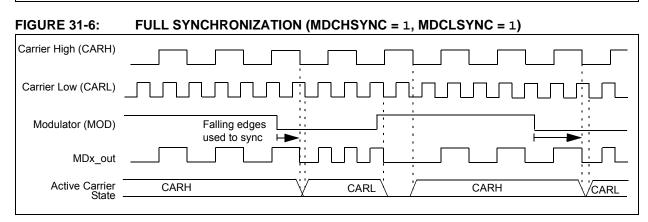


FIGURE 31-5:	CARRIER LOW SYNCHRONIZATION (MDCHSYNC = 0, MDCLSYNC = 1)
Carrier High (CARH)	
Carrier Low (CARL)	
Modulator (MOD)	
MDx_out	
Active Carrier State -	



31.5 Input and Output Through Pins

The modulation and carrier sources may be selected to come from any device pin with the PPS control logic. Selecting a pin requires two settings: The source selection determines that the PPS will be used and the PPS control selects the desired pin. Source and PPS registers are identified in Table 31-2. PPS register pin selections are shown in Register 12-1 and Register 12-2.

TABLE 31-2:

Source	Source Register	PPS Register
Modulation	MDxSRC	MDxMODPPS
Carrier High	MDxCARH	MDxCHPPS
Carrier Low	MDxCARL	MDxCLPPS

Any device pin can be selected as the modulation output with the individual pin PPS controls. See Register 12-2 for the pin output selections.

31.6 Carrier Source Polarity Select

The signal provided from any selected input source for the carrier high and carrier low signals can be inverted. Inverting the signal for the carrier high source is enabled by setting the CHPOL bit of the MDxCON1 register. Inverting the signal for the carrier low source is enabled by setting the CLPOL bit of the MDxCON1 register.

31.7 Programmable Modulator Data

The BIT bit of the MDxCON0 register can be selected as the source for the modulator signal. When the BIT source is selected then software generates the modulation signal by setting and clearing the BIT bit at the respective desired modulation high and low times.

31.8 Modulated Output Polarity

The modulated output signal provided on the MDxOUT pin can also be inverted. Inverting the modulated output signal is enabled by setting the OPOL bit of the MDxCON0 register.

31.9 Operation in Sleep Mode

The DSM module is not affected by Sleep mode. The DSM will operate during Sleep provided that the Carrier and Modulator input sources are also active during Sleep.

31.10 Effects of a Reset

Upon any device Reset, the data signal modulator module is disabled. The user's firmware is responsible for initializing the module before enabling the output. The registers are reset to their default values.

31.11 Register Definitions: Data Signal Modulator

Long bit name prefixes for the 10-bit DAC peripherals are shown in Table 31-3. Refer to **Section 1.1** "**Register and Bit naming conventions**" for more information

TABLE 31-3:

Peripheral	Bit Name Prefix
DSM1	DSM1
DSM2	DSM2
DSM3	DSM3
DSM4 ⁽¹⁾	DSM4

Note 1: PIC16(L)F1777/9 only.

REGISTER 31-1: MDxCON0: MODULATION CONTROL REGISTER 0

R/W-0/0	U-0	R-0/0	R/W-0/0	U-0	U-0	U-0	R/W-0/0
EN	—	OUT	OPOL	—	—	—	BIT
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7	EN: Modulator Module Enable bit
	 1 = Modulator module is enabled and mixing input signals 0 = Modulator module is disabled and has no output
bit 6	Unimplemented: Read as '0'
bit 5	OUT: Modulator Output bit
	Displays the current output value of the modulator module. ⁽¹⁾
bit 4	OPOL: Modulator Output Polarity Select bit
	1 = Modulator output signal is inverted. Idle high output.
	0 = Modulator output signal is not inverted. Idle low output.
bit 3-1	Unimplemented: Read as '0'
bit 0	BIT: Allows direct software control of the modulation source input to module ⁽²⁾
	1 = Modulator uses High Carrier source
	0 = Modulator uses Low Carrier source
	The modulated and whether many and he many the and an under some from the algorithm that whether the

- **Note 1:** The modulated output frequency can be greater and asynchronous from the clock that updates this register bit, the bit value may not be valid for higher speed modulator or carrier signals.
 - 2: BIT must be selected as the modulation source in the MDSRC register for this operation.

U-0	U-0	R/W-0/0	R/W-0/0	U-0	U-0	R/W-0/0	R/W-0/0	
_	_	CHPOL	CHSYNC	_	—	CLPOL	CLSYNC	
bit 7							bit 0	
Legend:								
R = Readab	ole bit	W = Writable	bit	U = Unimpler	mented bit, read	as '0'		
u = Bit is unchanged x = Bit is unknown -n/n = Value at POR and BOR/Value at all other R							ther Resets	
'1' = Bit is set '0' = Bit is cleared								
bit 7-6	Unimplemen	ted: Read as '	0'					
bit 5	CHPOL: Mod	lulation High C	arrier Polarity	Select bit				
	1 = Selected	high carrier so	ource is inverte	ed				
	0 = Selected	high carrier so	ource is not inv	verted				
bit 4	CHSYNC: Mo	odulation High	Carrier Synch	ronization Ena	ble bit			
	1 = Modulator waits for a low edge on the high carrier before allowing a switch to the low carr						ow carrier	
		0 = Modulator output is not synchronized to the high carrier ⁽¹⁾						
bit 3-2	Unimplemen	Unimplemented: Read as '0'						
bit 1	CLPOL: Mod	CLPOL: Modulation Low Carrier Polarity Select bit						
	1 = Selected	low carrier sou	urce is inverte	d				
	0 = Selected	low carrier sou	urce is not inv	erted				
bit 0	CLSYNC: Mo	odulation Low 0	Carrier Synchr	onization Enab	ole bit			
					fore allowing a	switch to the hi	gh carrier	
	0 = Modulato	or output is not	synchronized	to the low carr	ier ⁽¹⁾			

REGISTER 31-2: MDxCON1: MODULATION CONTROL REGISTER 1

Note 1: Narrowed carrier pulse widths or spurs may occur in the signal stream if the carrier is not synchronized.

REGISTER 31-3: MDxSRC: MODULATION SOURCE CONTROL REGISTER

U-0	U-0	U-0	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
—	—				MS<4:0>		
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-5 Unimplemented: Read as '0'

bit 4-0 **MS<4:0>** Modulation Source Selection bits See Table 31-4 or Table 31-5.

MS<4:0>	Modulation Source PIC16(L)F1777/9
11111	Reserved
11110	Reserved
11101	Reserved
11100	Reserved
11011	Reserved
11010	sync_C8OUT
11001	sync_C7OUT
11000	sync_C4OUT
10111	sync_C3OUT
10110	sync_C2OUT
10101	sync_C1OUT
10100	LC4_out
10011	LC3_out
10010	LC2_out
10001	LC1_out
10000	PWM12_out
01111	PWM11_out
01110	PWM6_out
01101	PWM5_out
01100	PWM10_out
01011	PWM9_out
01010	PWM4_out
01001	PWM3_out
01000	PWM8_out
00111	CCP7_out
00110	CCP2_out
00101	CCP1_out
00100	SDO_OUT
00011	DT
00010	TX_out
00001	MDxBIT
00000	MDxMODPPS pin selection

TABLE 31-4: MODULATION SOURCE

TABLE 31-5: MODULATION SOURCE

MS<4:0>	Modulation Source PIC16(L)F1778
11111	Reserved
11110	Reserved
11101	Reserved
11100	Reserved
11011	Reserved
11010	sync_C6OUT
11001	sync_C5OUT
11000	sync_C4OUT
10111	sync_C3OUT
10110	sync_C2OUT
10101	sync_C1OUT
10100	LC4_out
10011	LC3_out
10010	LC2_out
10001	LC1_out
10000	Reserved
01111	PWM11_out
01110	PWM6_out
01101	PWM5_out
01100	Reserved
01011	PWM9_out
01010	PWM4_out
01001	PWM3_out
01000	Reserved
00111	CCP7_out
00110	CCP2_out
00101	CCP1_out
00100	SDO_OUT
00011	DT
00010	TX_out
00001	MDxBIT
00000	MDxMODPPS pin selection

U-0	U-0	U-0	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	
—	_	—			CH<4:0>			
bit 7							bit 0	
Legend:								
R = Readable	bit	W = Writable	bit	U = Unimplen	nented bit, read	l as '0'		
u = Bit is unch	u = Bit is unchanged x = Bit is unknown		nown	-n/n = Value at POR and BOR/Value at all other Resets				
'1' = Bit is set		'0' = Bit is clea	ared					
'1' = Bit is set	-	'0' = Bit is clea	ared					

REGISTER 31-4: MDxCARH: MODULATION HIGH CARRIER CONTROL REGISTER

bit 7-5 Unimplemented: Read as '0'

bit 4-0 CH<4:0> Modulator Data High Carrier Selection bits⁽¹⁾ See Table 31-6.

Note 1: Narrowed carrier pulse widths or spurs may occur in the signal stream if the carrier is not synchronized.

CH<4:0>	Carrier Source	Carrier Source		
	PIC16(L)F1778	PIC16(L)F1777/9		
11111	Reserved	Reserved		
11110	Reserved	Reserved		
11101	Reserved	Reserved		
11100	Reserved	Reserved		
11011	Reserved	Reserved		
11010	Reserved	Reserved		
11001	Reserved	Reserved		
11000	Reserved	Reserved		
10111	Reserved	Reserved		
10110	Reserved	Reserved		
10101	Reserved	Reserved		
10100	Reserved	Reserved		
10011	Reserved	Reserved		
10010	LC4_out	LC4_out		
10001	LC3_out	LC3_out		
10000	LC2_out	LC2_out		
01111	LC1_out	LC1_out		
01110	Reserved	PWM12_out		
01101	PWM11_out	PWM11_out		
01100	PWM6_out	PWM6_out		
01011	PWM5_out	PWM5_out		
01010	Reserved	PWM10_out		
01001	PWM9_out	PWM9_out		
01000	PWM4_out	PWM4_out		
00111	PWM3_out	PWM3_out		
00110	Reserved	CCP8_out		
00101	CCP7_out	CCP7_out		
00100	CCP2_out	CCP2_out		
00011	CCP1_out	CCP1_out		
00010	HFINTOSC	HFINTOSC		
00001	FOSC	FOSC		
00000	MDxMODPPS pin selection	MDxMODPPS pin selection		

TABLE 31-6:HIGH CARRIER SOURCE

U-0	U-0	U-0	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	
—	_	—			CL<4:0>			
bit 7							bit 0	
Legend:								
R = Readable bit W = Writable bit		bit	U = Unimplemented bit, read as '0'					
u = Bit is unchanged x = Bit is unknown		nown	-n/n = Value a	at POR and BO	R/Value at all o	ther Resets		
'1' = Bit is set		'0' = Bit is clea	ared					

REGISTER 31-5: MDxCARL: MODULATION LOW CARRIER CONTROL REGISTER

bit 7-5 Unimplemented: Read as '0'

bit 4-0 CL<4:0> Modulator Data Low Carrier Selection bits⁽¹⁾ See Table 31-7.

Note 1: Narrowed carrier pulse widths or spurs may occur in the signal stream if the carrier is not synchronized.

TADLE 31-7:	LOW CARRIER SOURCE				
CH<4:0>	Carrier Source PIC16(L)F1778	Carrier Source PIC16(L)F1777/9			
11111	Reserved	Reserved			
11110	Reserved	Reserved			
11101	Reserved	Reserved			
11100	Reserved	Reserved			
11011	Reserved	Reserved			
11010	Reserved	Reserved			
11001	Reserved	Reserved			
11000	Reserved	Reserved			
10111	Reserved	Reserved			
10110	Reserved	Reserved			
10101	Reserved	Reserved			
10100	Reserved	Reserved			
10011	Reserved	Reserved			
10010	LC4_out	LC4_out			
10001	LC3_out	LC3_out			
10000	LC2_out	LC2_out			
01111	LC1_out	LC1_out			
01110	Reserved	PWM12_out			
01101	PWM11_out	PWM11_out			
01100	PWM6_out PWM6_out				
01011	PWM5_out PWM5_out				
01010	Reserved	PWM10_out			
01001	PWM9_out PWM9_out				
01000	PWM4_out	PWM4_out			
00111	PWM3_out	PWM3_out			
00110	Reserved	CCP8_out			
00101	CCP7_out	CCP7_out			
00100	CCP2_out	CCP2_out			
00011	CCP1_out	CCP1_out			
00010	HFINTOSC	HFINTOSC			
00001	FOSC	FOSC			
00000	MDxMODPPS pin selection MDxMODPPS pin selection				

TABLE 31-7: LOW CARRIER SOURCE

TABLE 31-8: SUMMARY OF REGISTERS ASSOCIATED WITH DATA SIGNAL MODULATOR MODE⁽¹⁾

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
MDxCARH	—	—	_			CH<4:0>			436
MDxCARL	—	_	_			CL<4:0>			438
MDxSRC	—	_	_	MS<4:0>				434	
MDxCON0	EN	_	OUT	OPOL	—	—	—	BIT	433
MDxCON1	_		CHPOL	CHSYNC	—	—	CLPOL	CLSYNC	433

Legend: — = unimplemented, read as '0'. Shaded cells are not used in the Data Signal Modulator mode.

Note 1: DSM4 available on PIC16LF1777/9 only.

32.0 MASTER SYNCHRONOUS SERIAL PORT (MSSP) MODULE

32.1 MSSP Module Overview

The Master Synchronous Serial Port (MSSP) module is a serial interface useful for communicating with other peripheral or microcontroller devices. These peripheral devices may be serial EEPROMs, shift registers, display drivers, A/D converters, etc. The MSSP module can operate in one of two modes:

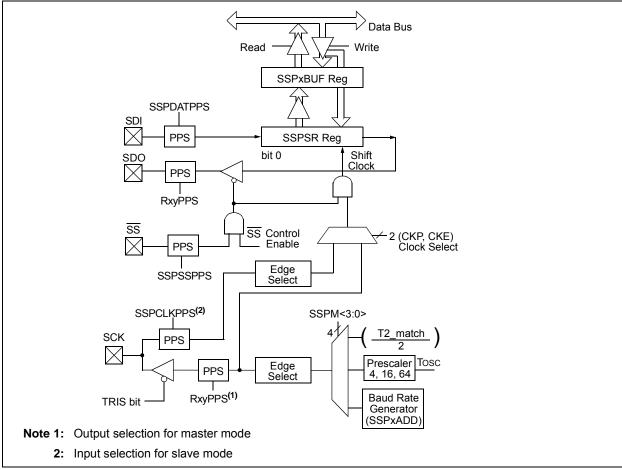
- Serial Peripheral Interface (SPI)
- Inter-Integrated Circuit (I²C)

The SPI interface supports the following modes and features:

- Master mode
- Slave mode
- Clock Parity
- Slave Select Synchronization (Slave mode only)
- · Daisy-chain connection of slave devices

Figure 32-1 is a block diagram of the SPI interface module.



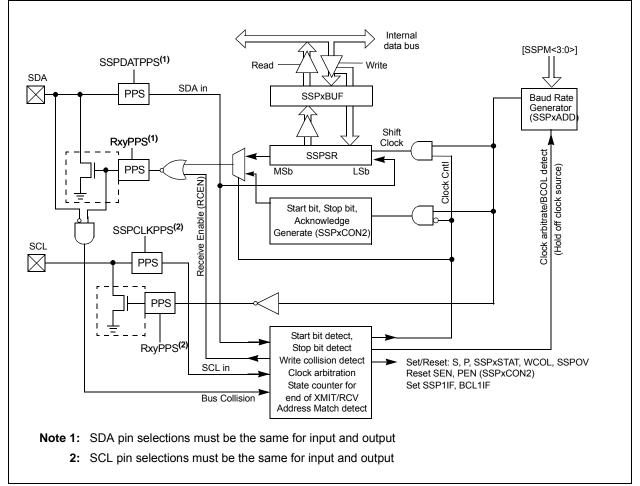


The I^2C interface supports the following modes and features:

- Master mode
- Slave mode
- Byte NACKing (Slave mode)
- · Limited multi-master support
- 7-bit and 10-bit addressing
- · Start and Stop interrupts
- · Interrupt masking
- Clock stretching
- · Bus collision detection
- · General call address matching
- Address masking
- Address Hold and Data Hold modes
- Selectable SDA hold times

Figure 32-2 is a block diagram of the I^2C interface module in Master mode. Figure 32-3 is a diagram of the I^2C interface module in Slave mode.





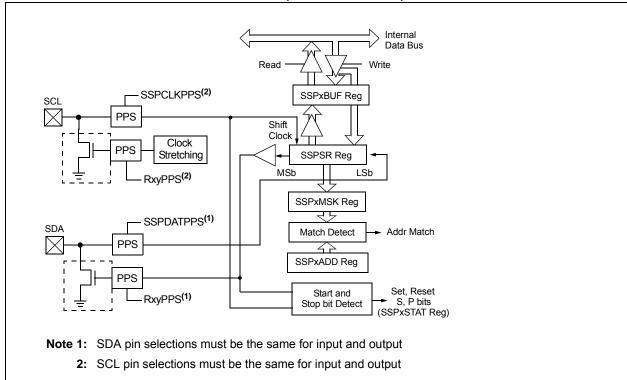


FIGURE 32-3: MSSP BLOCK DIAGRAM (I²C SLAVE MODE)

32.2 SPI Mode Overview

The Serial Peripheral Interface (SPI) bus is a synchronous serial data communication bus that operates in Full-Duplex mode. Devices communicate in a master/slave environment where the master device initiates the communication. A slave device is controlled through a Chip Select known as Slave Select.

The SPI bus specifies four signal connections:

- Serial Clock (SCK)
- Serial Data Out (SDO)
- Serial Data In (SDI)
- Slave Select (SS)

Figure 32-1 shows the block diagram of the MSSP module when operating in SPI mode.

The SPI bus operates with a single master device and one or more slave devices. When multiple slave devices are used, an independent Slave Select connection is required from the master device to each slave device.

Figure 32-4 shows a typical connection between a master device and multiple slave devices.

The master selects only one slave at a time. Most slave devices have tri-state outputs so their output signal appears disconnected from the bus when they are not selected.

Transmissions involve two shift registers, eight bits in size, one in the master and one in the slave. With either the master or the slave device, data is always shifted out one bit at a time, with the Most Significant bit (MSb) shifted out first. At the same time, a new Least Significant bit (LSb) is shifted into the same register.

Figure 32-5 shows a typical connection between two processors configured as master and slave devices.

Data is shifted out of both shift registers on the programmed clock edge and latched on the opposite edge of the clock.

The master device transmits information out on its SDO output pin which is connected to, and received by, the slave's SDI input pin. The slave device transmits information out on its SDO output pin, which is connected to, and received by, the master's SDI input pin.

To begin communication, the master device first sends out the clock signal. Both the master and the slave devices should be configured for the same clock polarity.

The master device starts a transmission by sending out the MSb from its shift register. The slave device reads this bit from that same line and saves it into the LSb position of its shift register. During each SPI clock cycle, a full-duplex data transmission occurs. This means that while the master device is sending out the MSb from its shift register (on its SDO pin) and the slave device is reading this bit and saving it as the LSb of its shift register, that the slave device is also sending out the MSb from its shift register (on its SDO pin) and the master device is reading this bit and saving it as the LSb of its shift register.

After eight bits have been shifted out, the master and slave have exchanged register values.

If there is more data to exchange, the shift registers are loaded with new data and the process repeats itself.

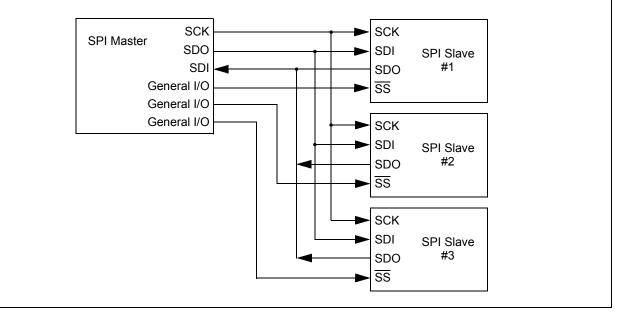
Whether the data is meaningful or not (dummy data), depends on the application software. This leads to three scenarios for data transmission:

- Master sends useful data and slave sends dummy data.
- Master sends useful data and slave sends useful data.
- Master sends dummy data and slave sends useful data.

Transmissions may involve any number of clock cycles. When there is no more data to be transmitted, the master stops sending the clock signal and it deselects the slave.

Every slave device connected to the bus that has not been selected through its slave select line must disregard the clock and transmission signals and must not transmit out any data of its own.

FIGURE 32-4: SPI MASTER AND MULTIPLE SLAVE CONNECTION



32.2.1 SPI MODE REGISTERS

The MSSP module has five registers for SPI mode operation. These are:

- MSSP STATUS register (SSPxSTAT)
- MSSP Control register 1 (SSPxCON1)
- MSSP Control register 3 (SSPxCON3)
- MSSP Data Buffer register (SSPxBUF)
- MSSP Address register (SSPxADD)
- MSSP Shift register (SSPSR) (Not directly accessible)

SSPxCON1 and SSPxSTAT are the control and STATUS registers in SPI mode operation. The SSPxCON1 register is readable and writable. The lower six bits of the SSPxSTAT are read-only. The upper two bits of the SSPxSTAT are read/write.

In one SPI Master mode, SSPxADD can be loaded with a value used in the Baud Rate Generator. More information on the Baud Rate Generator is available in Section 32.7 "Baud Rate Generator".

SSPSR is the shift register used for shifting data in and out. SSPxBUF provides indirect access to the SSPSR register. SSPxBUF is the buffer register to which data bytes are written, and from which data bytes are read.

In receive operations, SSPSR and SSPxBUF together create a buffered receiver. When SSPSR receives a complete byte, it is transferred to SSPxBUF and the SSPxIF interrupt is set.

During transmission, the SSPxBUF is not buffered. A write to SSPxBUF will write to both SSPxBUF and SSPSR.

32.2.2 SPI MODE OPERATION

When initializing the SPI, several options need to be specified. This is done by programming the appropriate control bits (SSPxCON1<5:0> and SSPxSTAT<7:6>). These control bits allow the following to be specified:

- Master mode (SCK is the clock output)
- Slave mode (SCK is the clock input)
- · Clock Polarity (Idle state of SCK)
- Data Input Sample Phase (middle or end of data output time)
- Clock Edge (output data on rising/falling edge of SCK)
- Clock Rate (Master mode only)
- Slave Select mode (Slave mode only)

To enable the serial port, SSP Enable bit, SSPEN of the SSPxCON1 register, must be set. To reset or reconfigure SPI mode, clear the SSPEN bit, re-initialize the SSPxCONx registers and then set the SSPEN bit. This configures the SDI, SDO, SCK and \overline{SS} pins as serial port pins. For the pins to behave as the serial port function, some must have their data direction bits (in the TRIS register) appropriately programmed as follows:

- · SDI must have corresponding TRIS bit set
- · SDO must have corresponding TRIS bit cleared
- SCK (Master mode) must have corresponding
 TRIS bit cleared
- SCK (Slave mode) must have corresponding
 TRIS bit set
- SS must have corresponding TRIS bit set

Any serial port function that is not desired may be overridden by programming the corresponding data direction (TRIS) register to the opposite value. The MSSP consists of a transmit/receive shift register (SSPSR) and a buffer register (SSPxBUF). The SSPSR shifts the data in and out of the device, MSb first. The SSPxBUF holds the data that was written to the SSPSR until the received data is ready. Once the eight bits of data have been received, that byte is moved to the SSPxBUF register. Then, the Buffer Full Detect bit, BF of the SSPxSTAT register, and the interrupt flag bit, SSPxIF, are set. This double-buffering of the received data (SSPxBUF) allows the next byte to start reception before reading the data that was just received. Any write to the SSPxBUF register during transmission/reception of data will be ignored and the Write Collision Detect bit WCOL of the SSPxCON1 register, will be set. User software must clear the WCOL bit to allow the following write(s) to the SSPxBUF register to complete successfully.

When the application software is expecting to receive valid data, the SSPxBUF should be read before the next byte of data to transfer is written to the SSPxBUF. The Buffer Full bit, BF of the SSPxSTAT register, indicates when SSPxBUF has been loaded with the received data (transmission is complete). When the SSPxBUF is read, the BF bit is cleared. This data may be irrelevant if the SPI is only a transmitter. Generally, the MSSP interrupt is used to determine when the transmission/reception has completed. If the interrupt method is not going to be used, then software polling can be done to ensure that a write collision does not occur.

The SSPSR is not directly readable or writable and can only be accessed by addressing the SSPxBUF register. Additionally, the SSPxSTAT register indicates the various Status conditions.

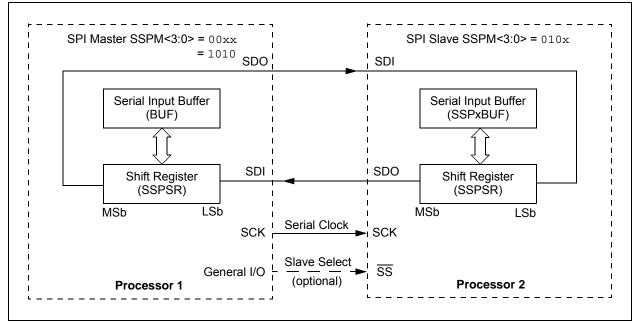


FIGURE 32-5: SPI MASTER/SLAVE CONNECTION

32.2.3 SPI MASTER MODE

The master can initiate the data transfer at any time because it controls the SCK line. The master determines when the slave (Processor 2, Figure 32-5) is to broadcast data by the software protocol.

In Master mode, the data is transmitted/received as soon as the SSPxBUF register is written to. If the SPI is only going to receive, the SDO output could be disabled (programmed as an input). The SSPSR register will continue to shift in the signal present on the SDI pin at the programmed clock rate. As each byte is received, it will be loaded into the SSPxBUF register as if a normal received byte (interrupts and Status bits appropriately set). The clock polarity is selected by appropriately programming the CKP bit of the SSPxCON1 register and the CKE bit of the SSPxSTAT register. This then, would give waveforms for SPI communication as shown in Figure 32-6, Figure 32-8, Figure 32-9 and Figure 32-10, where the MSB is transmitted first. In Master mode, the SPI clock rate (bit rate) is user programmable to be one of the following:

- Fosc/4 (or Tcy)
- Fosc/16 (or 4 * Tcy)
- Fosc/64 (or 16 * Tcy)
- Timer2 output/2
- Fosc/(4 * (SSPxADD + 1))

Figure 32-6 shows the waveforms for Master mode.

When the CKE bit is set, the SDO data is valid before there is a clock edge on SCK. The change of the input sample is shown based on the state of the SMP bit. The time when the SSPxBUF is loaded with the received data is shown.

Note: In Master mode the clock signal output to the SCK pin is also the clock signal input to the peripheral. The pin selected for output with the RxyPPS register must also be selected as the peripheral input with the SSPCLKPPS register.

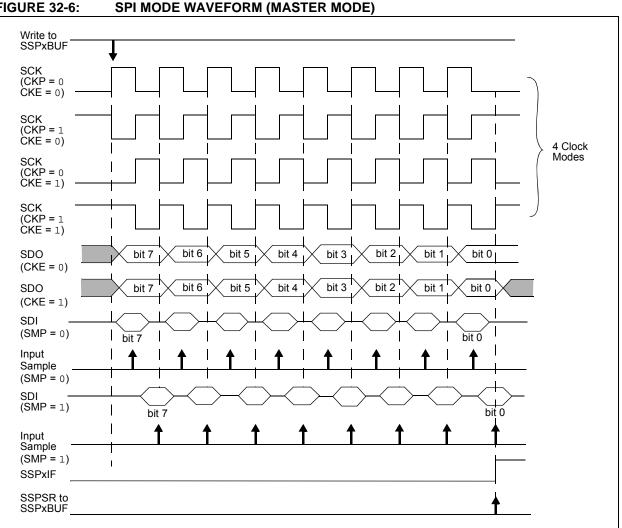


FIGURE 32-6:

32.2.4 SPI SLAVE MODE

In Slave mode, the data is transmitted and received as external clock pulses appear on SCK. When the last bit is latched, the SSPxIF interrupt flag bit is set.

Before enabling the module in SPI Slave mode, the clock line must match the proper Idle state. The clock line can be observed by reading the SCK pin. The Idle state is determined by the CKP bit of the SSPxCON1 register.

While in Slave mode, the external clock is supplied by the external clock source on the SCK pin. This external clock must meet the minimum high and low times as specified in the electrical specifications.

While in Sleep mode, the slave can transmit/receive data. The shift register is clocked from the SCK pin input and when a byte is received, the device will generate an interrupt. If enabled, the device will wake-up from Sleep.

32.2.4.1 Daisy-Chain Configuration

The SPI bus can sometimes be connected in a daisy-chain configuration. The first slave output is connected to the second slave input, the second slave output is connected to the third slave input, and so on. The final slave output is connected to the master input. Each slave sends out, during a second group of clock pulses, an exact copy of what was received during the first group of clock pulses. The whole chain acts as one large communication shift register. The daisy-chain feature only requires a single Slave Select line from the master device.

Figure 32-7 shows the block diagram of a typical daisy-chain connection when operating in SPI mode.

In a daisy-chain configuration, only the most recent byte on the bus is required by the slave. Setting the BOEN bit of the SSPxCON3 register will enable writes to the SSPxBUF register, even if the previous byte has not been read. This allows the software to ignore data that may not apply to it.

32.2.5 SLAVE SELECT SYNCHRONIZATION

The Slave Select can also be used to synchronize communication. The Slave Select line is held high until the master device is ready to communicate. When the Slave Select line is pulled low, the slave knows that a new transmission is starting.

If the slave fails to receive the communication properly, it will be reset at the end of the transmission, when the Slave Select line returns to a high state. The slave is then ready to receive a new transmission when the Slave Select line is pulled low again. If the Slave Select line is not used, there is a risk that the slave will eventually become out of sync with the master. If the slave misses a bit, it will always be one bit off in future transmissions. Use of the Slave Select line allows the slave and master to align themselves at the beginning of each transmission.

The \overline{SS} pin allows a Synchronous Slave mode. The SPI must be in Slave mode with \overline{SS} pin control enabled (SSPxCON1<3:0> = 0100).

When the \overline{SS} pin is low, transmission and reception are enabled and the SDO pin is driven.

When the \overline{SS} pin goes high, the SDO pin is no longer driven, even if in the middle of a transmitted byte and becomes a floating output. External pull-up/pull-down resistors may be desirable depending on the application.

Note 1:	When the SPI is in Slave mode with \overline{SS} pin				
	control enabled (SSPxCON1<3:0> =				
	0100), the SPI module will reset if the SS				
	pin is set to VDD.				
2:	When the SPI is used in Slave mode with				
	CKE set; the user must enable \overline{SS} pin				

control. **3:** While operated in SPI Slave mode the SMP bit of the SSPxSTAT register must remain clear.

When the SPI module resets, the bit counter is forced to '0'. This can be done by either forcing the SS pin to a high level or clearing the SSPEN bit.

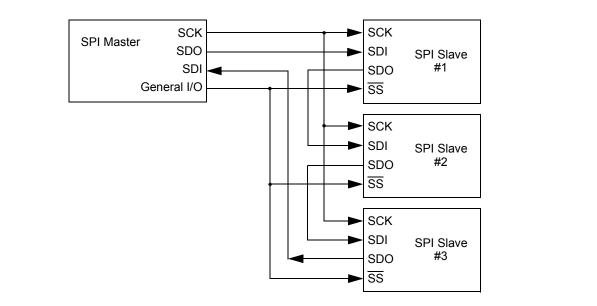
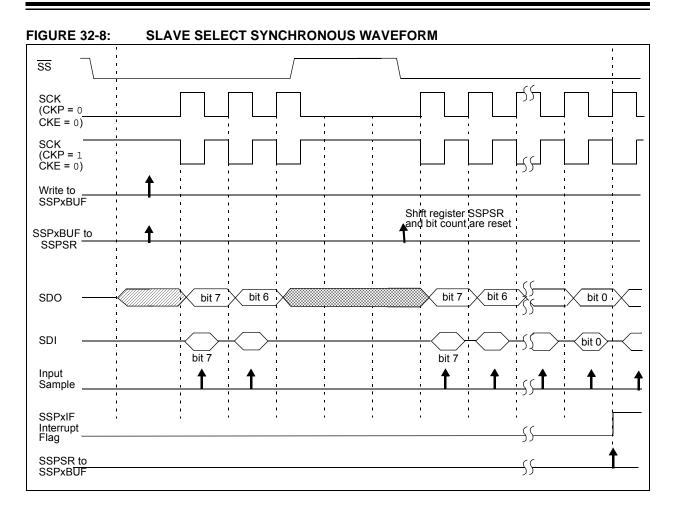


FIGURE 32-7: SPI DAISY-CHAIN CONNECTION



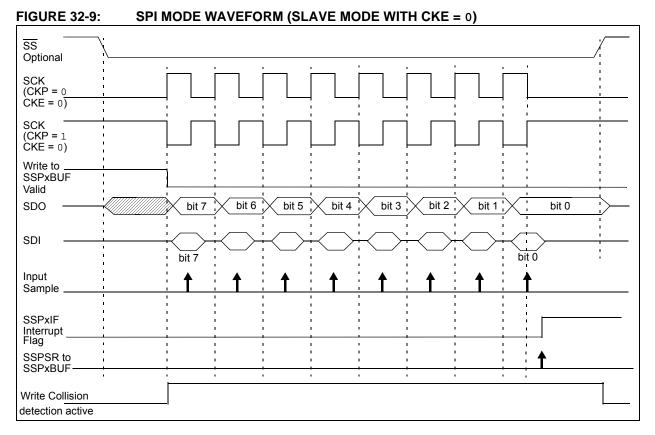
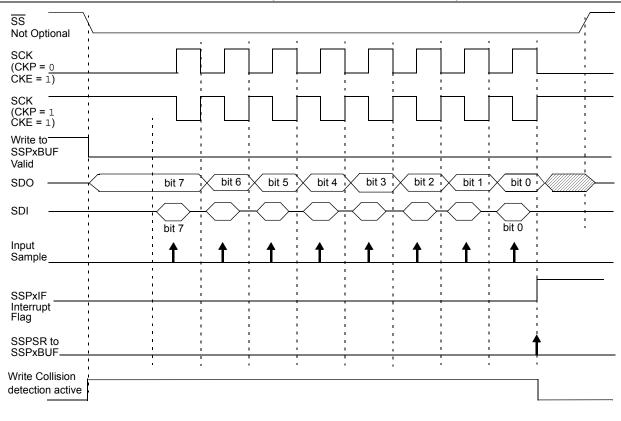


FIGURE 32-10: SPI MODE WAVEFORM (SLAVE MODE WITH CKE = 1)



32.2.6 SPI OPERATION IN SLEEP MODE

In SPI Master mode, module clocks may be operating at a different speed than when in Full-Power mode; in the case of the Sleep mode, all clocks are halted.

Special care must be taken by the user when the MSSP clock is much faster than the system clock.

In Slave mode, when MSSP interrupts are enabled, after the master completes sending data, an MSSP interrupt will wake the controller from Sleep.

If an exit from Sleep mode is not desired, MSSP interrupts should be disabled.

In SPI Master mode, when the Sleep mode is selected, all module clocks are halted and the transmission/reception will remain in that state until the device wakes. After the device returns to Run mode, the module will resume transmitting and receiving data.

In SPI Slave mode, the SPI Transmit/Receive Shift register operates asynchronously to the device. This allows the device to be placed in Sleep mode and data to be shifted into the SPI Transmit/Receive Shift register. When all eight bits have been received, the MSSP interrupt flag bit will be set and if enabled, will wake the device.

TABLE 32-1. SUMMART OF REGISTERS ASSOCIATED WITH SITCH ERATION									
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ANSELA	—		ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0	177
ANSELC	ANSC7	ANSC6	ANSC5	ANSC4	ANSC3	ANSC2	_	_	187
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	132
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	CCP1IE	TMR2IE	TMR1IE	133
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	CCP1IF	TMR2IF	TMR1IF	139
RxyPPS	—	_		RxyPPS<5:0>					205
SSPCLKPPS	—	_		SSPCLKPPS<5:0>					205, 207
SSPDATPPS	—	_	SSPDATPPS<5:0>					205, 207	
SSPSSPPS	—	_	SSPSSPPS<5:0>					205, 207	
SSP1BUF	Synchronous Serial Port Receive Buffer/Transmit Register						444*		
SSP1CON1	WCOL	SSPOV	SSPEN CKP SSPM<3:0>					489	
SSP1CON3	ACKTIM	PCIE	SCIE	BOEN	SDAHT	SBCDE	AHEN	DHEN	488
SSP1STAT	SMP	CKE	D/A	Р	S	R/W	UA	BF	488
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	176
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	181
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	186

TABLE 32-1: SUMMARY OF REGISTERS ASSOCIATED WITH SPI OPERATION

Legend: — = Unimplemented location, read as '0'. Shaded cells are not used by the MSSP in SPI mode.

Page provides register information.

32.3 I²C MODE OVERVIEW

The Inter-Integrated Circuit (I^2C) bus is a multi-master serial data communication bus. Devices communicate in a master/slave environment where the master devices initiate the communication. A slave device is controlled through addressing.

The I²C bus specifies two signal connections:

- · Serial Clock (SCL)
- · Serial Data (SDA)

Figure 32-11 shows the block diagram of the MSSP module when operating in I^2C mode.

Both the SCL and SDA connections are bidirectional open-drain lines, each requiring pull-up resistors for the supply voltage. Pulling the line to ground is considered a logical zero and letting the line float is considered a logical one.

Figure 32-11 shows a typical connection between two processors configured as master and slave devices.

The I^2C bus can operate with one or more master devices and one or more slave devices.

There are four potential modes of operation for a given device:

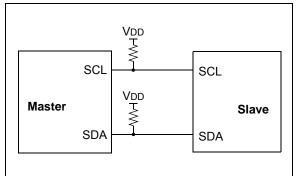
- Master Transmit mode
 (master is transmitting data to a slave)
- Master Receive mode
 (master is receiving data from a slave)
- Slave Transmit mode (slave is transmitting data to a master)
- Slave Receive mode (slave is receiving data from the master)

To begin communication, a master device starts out in Master Transmit mode. The master device sends out a Start bit followed by the address byte of the slave it intends to communicate with. This is followed by a single Read/Write bit, which determines whether the master intends to transmit to or receive data from the slave device.

If the requested slave exists on the bus, it will respond with an Acknowledge bit, otherwise known as an ACK. The master then continues in either Transmit mode or Receive mode and the slave continues in the complement, either in Receive mode or Transmit mode, respectively.

A Start bit is indicated by a high-to-low transition of the SDA line while the SCL line is held high. Address and data bytes are sent out, Most Significant bit (MSb) first. The Read/Write bit is sent out as a logical one when the master intends to read data from the slave, and is sent out as a logical zero when it intends to write data to the slave.

FIGURE 32-11: I²C MASTER/ SLAVE CONNECTION



The Acknowledge bit (\overline{ACK}) is an active-low signal, which holds the SDA line low to indicate to the transmitter that the slave device has received the transmitted data and is ready to receive more.

The transition of a data bit is always performed while the SCL line is held low. Transitions that occur while the SCL line is held high are used to indicate Start and Stop bits.

If the master intends to write to the slave, then it repeatedly sends out a byte of data, with the slave responding after each byte with an \overline{ACK} bit. In this example, the master device is in Master Transmit mode and the slave is in Slave Receive mode.

If the master intends to read from the slave, then it repeatedly receives a byte of data from the slave, and responds after each byte with an ACK bit. In this example, the master device is in Master Receive mode and the slave is Slave Transmit mode.

On the last byte of data communicated, the master device may end the transmission by sending a Stop bit. If the master device is in Receive mode, it sends the Stop bit in place of the last ACK bit. A Stop bit is indicated by a low-to-high transition of the SDA line while the SCL line is held high.

In some cases, the master may want to maintain control of the bus and re-initiate another transmission. If so, the master device may send another Start bit in place of the Stop bit or last ACK bit when it is in receive mode.

The I²C bus specifies three message protocols:

- Single message where a master writes data to a slave.
- Single message where a master reads data from a slave.
- Combined message where a master initiates a minimum of two writes, or two reads, or a combination of writes and reads, to one or more slaves.

When one device is transmitting a logical one, or letting the line float, and a second device is transmitting a logical zero, or holding the line low, the first device can detect that the line is not a logical one. This detection, when used on the SCL line, is called clock stretching. Clock stretching gives slave devices a mechanism to control the flow of data. When this detection is used on the SDA line, it is called arbitration. Arbitration ensures that there is only one master device communicating at any single time.

32.3.1 CLOCK STRETCHING

When a slave device has not completed processing data, it can delay the transfer of more data through the process of clock stretching. An addressed slave device may hold the SCL clock line low after receiving or sending a bit, indicating that it is not yet ready to continue. The master that is communicating with the slave will attempt to raise the SCL line in order to transfer the next bit, but will detect that the clock line has not yet been released. Because the SCL connection is open-drain, the slave has the ability to hold that line low until it is ready to continue communicating.

Clock stretching allows receivers that cannot keep up with a transmitter to control the flow of incoming data.

32.3.2 ARBITRATION

Each master device must monitor the bus for Start and Stop bits. If the device detects that the bus is busy, it cannot begin a new message until the bus returns to an Idle state.

However, two master devices may try to initiate a transmission on or about the same time. When this occurs, the process of arbitration begins. Each transmitter checks the level of the SDA data line and compares it to the level that it expects to find. The first transmitter to observe that the two levels do not match, loses arbitration, and must stop transmitting on the SDA line.

For example, if one transmitter holds the SDA line to a logical one (lets it float) and a second transmitter holds it to a logical zero (pulls it low), the result is that the SDA line will be low. The first transmitter then observes that the level of the line is different than expected and concludes that another transmitter is communicating.

The first transmitter to notice this difference is the one that loses arbitration and must stop driving the SDA line. If this transmitter is also a master device, it also must stop driving the SCL line. It then can monitor the lines for a Stop condition before trying to reissue its transmission. In the meantime, the other device that has not noticed any difference between the expected and actual levels on the SDA line continues with its original transmission. It can do so without any complications, because so far, the transmission appears exactly as expected with no other transmitter disturbing the message.

Slave Transmit mode can also be arbitrated, when a master addresses multiple slaves, but this is less common.

If two master devices are sending a message to two different slave devices at the address stage, the master sending the lower slave address always wins arbitration. When two master devices send messages to the same slave address, and addresses can sometimes refer to multiple slaves, the arbitration process must continue into the data stage.

Arbitration usually occurs very rarely, but it is a necessary process for proper multi-master support.

32.4 I²C MODE OPERATION

All MSSP I²C communication is byte oriented and shifted out MSb first. Six SFR registers and two interrupt flags interface the module with the $PIC^{\mathbb{R}}$ microcontroller and user software. Two pins, SDA and SCL, are exercised by the module to communicate with other external I²C devices.

32.4.1 BYTE FORMAT

All communication in I^2C is done in 9-bit segments. A byte is sent from a master to a slave or vice-versa, followed by an Acknowledge bit sent back. After the eighth falling edge of the SCL line, the device outputting data on the SDA changes that pin to an input and reads in an acknowledge value on the next clock pulse.

The clock signal, SCL, is provided by the master. Data is valid to change while the SCL signal is low, and sampled on the rising edge of the clock. Changes on the SDA line while the SCL line is high define special conditions on the bus, explained below.

32.4.2 DEFINITION OF I²C TERMINOLOGY

There is language and terminology in the description of I^2C communication that have definitions specific to I^2C . That word usage is defined below and may be used in the rest of this document without explanation. This table was adapted from the Philips I^2C specification.

32.4.3 SDA AND SCL PINS

Selection of any I²C mode with the SSPEN bit set, forces the SCL and SDA pins to be open-drain. These pins should be set by the user to inputs by setting the appropriate TRIS bits.

- Note 1: Data is tied to output zero when an I²C mode is enabled.
 - 2: Any device pin can be selected for SDA and SCL functions with the PPS peripheral. These functions are bidirectional. The SDA input is selected with the SSPDATPPS registers. The SCL input is selected with the SSPCLKPPS registers. Outputs are selected with the RxyPPS registers. It is the user's responsibility to make the selections so that both the input and the output for each function is on the same pin.

32.4.4 SDA HOLD TIME

The hold time of the SDA pin is selected by the SDAHT bit of the SSPxCON3 register. Hold time is the time SDA is held valid after the falling edge of SCL. Setting the SDAHT bit selects a longer 300 ns minimum hold time and may help on buses with large capacitance.

TABLE 32-2:I²C BUS TERMS

TADLE 32-2.	I C BUS IERINIS
TERM	Description
Transmitter	The device which shifts data out onto the bus.
Receiver	The device which shifts data in from the bus.
Master	The device that initiates a transfer, generates clock signals and termi- nates a transfer.
Slave	The device addressed by the master.
Multi-master	A bus with more than one device that can initiate data transfers.
Arbitration	Procedure to ensure that only one master at a time controls the bus. Winning arbitration ensures that the message is not corrupted.
Synchronization	Procedure to synchronize the clocks of two or more devices on the bus.
Idle	No master is controlling the bus, and both SDA and SCL lines are high.
Active	Any time one or more master devices are controlling the bus.
Addressed Slave	Slave device that has received a matching address and is actively being clocked by a master.
Matching Address	Address byte that is clocked into a slave that matches the value stored in SSPxADD.
Write Request	Slave receives a matching address with R/W bit clear, and is ready to clock in data.
Read Request	Master sends an address byte with the R/W bit set, indicating that it wishes to clock data out of the Slave. This data is the next and all following bytes until a Restart or Stop.
Clock Stretching	When a device on the bus hold SCL low to stall communication.
Bus Collision	Any time the SDA line is sampled low by the module while it is out- putting and expected high state.

32.4.5 START CONDITION

The I^2C specification defines a Start condition as a transition of SDA from a high to a low state while SCL line is high. A Start condition is always generated by the master and signifies the transition of the bus from an Idle to an Active state. Figure 32-12 shows wave forms for Start and Stop conditions.

A bus collision can occur on a Start condition if the module samples the SDA line low before asserting it low. This does not conform to the I^2C Specification that states no bus collision can occur on a Start.

32.4.6 STOP CONDITION

A Stop condition is a transition of the SDA line from low-to-high state while the SCL line is high.

Note: At least one SCL low time must appear before a Stop is valid, therefore, if the SDA line goes low then high again while the SCL line stays high, only the Start condition is detected.

32.4.7 RESTART CONDITION

A Restart is valid any time that a Stop would be valid. A master can issue a Restart if it wishes to hold the bus after terminating the current transfer. A Restart has the same effect on the slave that a Start would, resetting all slave logic and preparing it to clock in an address. The master may want to address the same or another slave. Figure 32-13 shows the wave form for a Restart condition.

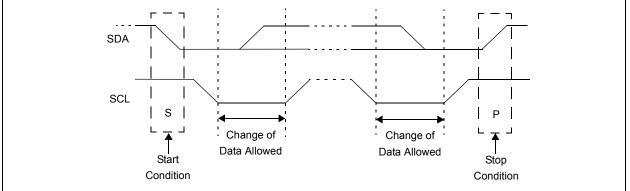
In 10-bit Addressing Slave mode a Restart is required for the master to clock data out of the addressed slave. Once a slave has been fully addressed, matching both high and low address bytes, the master can issue a Restart and the high address byte with the R/\overline{W} bit set. The slave logic will then hold the clock and prepare to clock out data.

After a full match with R/W clear in 10-bit mode, a prior match flag is set and maintained until a Stop condition, a high address with R/W clear, or high address match fails.

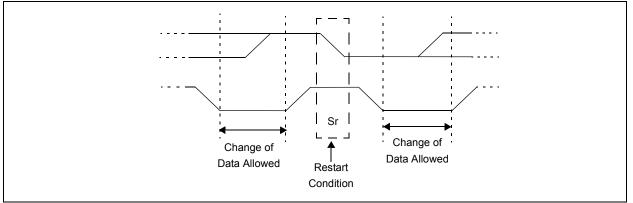
32.4.8 START/STOP CONDITION INTERRUPT MASKING

The SCIE and PCIE bits of the SSPxCON3 register can enable the generation of an interrupt in Slave modes that do not typically support this function. Slave modes where interrupt on Start and Stop detect are already enabled, these bits will have no effect.









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32.4.9 ACKNOWLEDGE SEQUENCE

The ninth SCL pulse for any transferred byte in I^2C is dedicated as an Acknowledge. It allows receiving devices to respond back to the transmitter by pulling the SDA line low. The transmitter must release control of the line during this time to shift in the response. The Acknowledge (ACK) is an active-low signal, pulling the SDA line low indicates to the transmitter that the device has received the transmitted data and is ready to receive more.

The result of an \overline{ACK} is placed in the ACKSTAT bit of the SSPxCON2 register.

Slave software, when the AHEN and DHEN bits are set, allow the user to set the ACK value sent back to the transmitter. The ACKDT bit of the SSPxCON2 register is set/cleared to determine the response.

Slave hardware will generate an ACK response if the AHEN and DHEN bits of the SSPxCON3 register are clear.

There are certain conditions where an \overline{ACK} will not be sent by the slave. If the BF bit of the SSPxSTAT register or the SSPOV bit of the SSPxCON1 register are set when a byte is received.

When the module is addressed, after the eighth falling edge of SCL on the bus, the ACKTIM bit of the SSPxCON3 register is set. The ACKTIM bit indicates the acknowledge time of the active bus. The ACKTIM Status bit is only active when the AHEN bit or DHEN bit is enabled.

32.5 I²C SLAVE MODE OPERATION

The MSSP Slave mode operates in one of four modes selected by the SSPM bits of SSPxCON1 register. The modes can be divided into 7-bit and 10-bit Addressing mode. 10-bit Addressing modes operate the same as 7-bit with some additional overhead for handling the larger addresses.

Modes with Start and Stop bit interrupts operate the same as the other modes with SSPxIF additionally getting set upon detection of a Start, Restart, or Stop condition.

32.5.1 SLAVE MODE ADDRESSES

The SSPxADD register (Register 32-6) contains the Slave mode address. The first byte received after a Start or Restart condition is compared against the value stored in this register. If the byte matches, the value is loaded into the SSPxBUF register and an interrupt is generated. If the value does not match, the module goes idle and no indication is given to the software that anything happened.

The SSP Mask register (Register 32-5) affects the address matching process. See Section 32.5.8 "SSP Mask Register" for more information.

32.5.1.1 I²C Slave 7-bit Addressing Mode

In 7-bit Addressing mode, the LSb of the received data byte is ignored when determining if there is an address match.

32.5.1.2 I²C Slave 10-bit Addressing Mode

In 10-bit Addressing mode, the first received byte is compared to the binary value of '1 1 1 1 0 A9 A8 0'. A9 and A8 are the two MSb's of the 10-bit address and stored in bits 2 and 1 of the SSPxADD register.

After the acknowledge of the high byte the UA bit is set and SCL is held low until the user updates SSPxADD with the low address. The low address byte is clocked in and all eight bits are compared to the low address value in SSPxADD. Even if there is not an address match; SSPxIF and UA are set, and SCL is held low until SSPxADD is updated to receive a high byte again. When SSPxADD is updated the UA bit is cleared. This ensures the module is ready to receive the high address byte on the next communication.

A high and low address match as a write request is required at the start of all 10-bit addressing communication. A transmission can be initiated by issuing a Restart once the slave is addressed, and clocking in the high address with the R/W bit set. The slave hardware will then acknowledge the read request and prepare to clock out data. This is only valid for a slave after it has received a complete high and low address byte match.

32.5.2 SLAVE RECEPTION

When the R/\overline{W} bit of a matching received address byte is clear, the R/\overline{W} bit of the SSPxSTAT register is cleared. The received address is loaded into the SSPxBUF register and acknowledged.

When the overflow condition exists for a received address, then not Acknowledge is given. An overflow condition is defined as either bit BF of the SSPxSTAT register is set, or bit SSPOV of the SSPxCON1 register is set. The BOEN bit of the SSPxCON3 register modifies this operation. For more information see Register 32-4.

An MSSP interrupt is generated for each transferred data byte. Flag bit, SSPxIF, must be cleared by software.

When the SEN bit of the SSPxCON2 register is set, SCL will be held low (clock stretch) following each received byte. The clock must be released by setting the CKP bit of the SSPxCON1 register, except sometimes in 10-bit mode. See Section 32.5.6.2 "10-bit Addressing Mode" for more detail.

32.5.2.1 7-bit Addressing Reception

This section describes a standard sequence of events for the MSSP module configured as an I^2C slave in 7-bit Addressing mode. Figure 32-14 and Figure 32-15 is used as a visual reference for this description.

This is a step by step process of what typically must be done to accomplish I²C communication.

- 1. Start bit detected.
- 2. S bit of SSPxSTAT is set; SSPxIF is set if interrupt on Start detect is enabled.
- 3. Matching address with R/\overline{W} bit clear is received.
- 4. The slave pulls SDA low sending an ACK to the master, and sets SSPxIF bit.
- 5. Software clears the SSPxIF bit.
- 6. Software reads received address from SSPxBUF clearing the BF flag.
- 7. If SEN = 1; Slave software sets CKP bit to release the SCL line.
- 8. The master clocks out a data byte.
- 9. Slave drives SDA low sending an ACK to the master, and sets SSPxIF bit.
- 10. Software clears SSPxIF.
- 11. Software reads the received byte from SSPxBUF clearing BF.
- 12. Steps 8-12 are repeated for all received bytes from the master.
- 13. Master sends Stop condition, setting P bit of SSPxSTAT, and the bus goes idle.

32.5.2.2 7-bit Reception with AHEN and DHEN

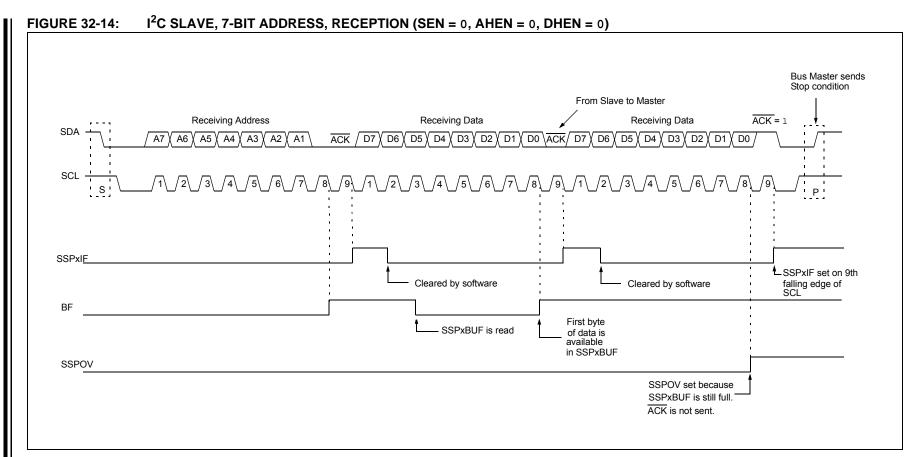
Slave device reception with AHEN and DHEN set operate the same as without these options with extra interrupts and clock stretching added after the eighth falling edge of SCL. These additional interrupts allow the slave software to decide whether it wants to ACK the receive address or data byte, rather than the hardware. This functionality adds support for PMBus[™] that was not present on previous versions of this module.

This list describes the steps that need to be taken by slave software to use these options for I^2C communication. Figure 32-16 displays a module using both address and data holding. Figure 32-17 includes the operation with the SEN bit of the SSPxCON2 register set.

- 1. S bit of SSPxSTAT is set; SSPxIF is set if interrupt on Start detect is enabled.
- Matching address with R/W bit clear is clocked in. SSPxIF is set and CKP cleared after the eighth falling edge of SCL.
- 3. Slave clears the SSPxIF.
- Slave can look at the ACKTIM bit of the SSPxCON3 register to <u>determine</u> if the SSPxIF was after or before the ACK.
- 5. Slave reads the address value from SSPxBUF, clearing the BF flag.
- 6. Slave sets ACK value clocked out to the master by setting ACKDT.
- 7. Slave releases the clock by setting CKP.
- 8. SSPxIF is set after an \overline{ACK} , not after a NACK.
- 9. If SEN = 1 the slave hardware will stretch the clock after the ACK.
- 10. Slave clears SSPxIF.

Note: SSPxIF is still set after the ninth falling edge of SCL even if there is no clock stretching and BF has been cleared. Only if NACK is sent to master is SSPxIF not set

- 11. SSPxIF set and CKP cleared after eighth falling edge of SCL for a received data byte.
- 12. Slave looks at ACKTIM bit of SSPxCON3 to determine the source of the interrupt.
- 13. Slave reads the received data from SSPxBUF clearing BF.
- 14. Steps 7-14 are the same for each received data byte.
- 15. Communication is ended by either the slave sending an ACK = 1, or the master sending a Stop condition. If a Stop is sent and interrupt on Stop detect is disabled, the slave will only know by polling the P bit of the SSTSTAT register.



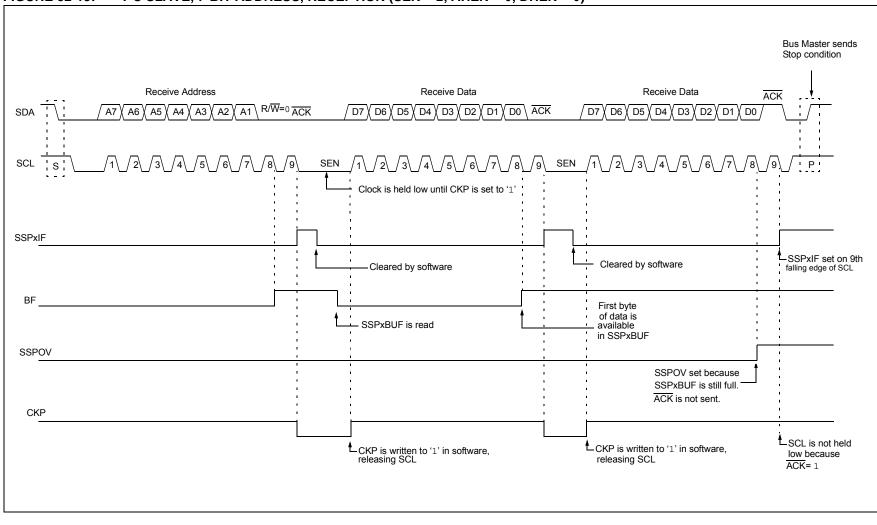
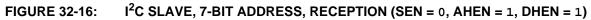
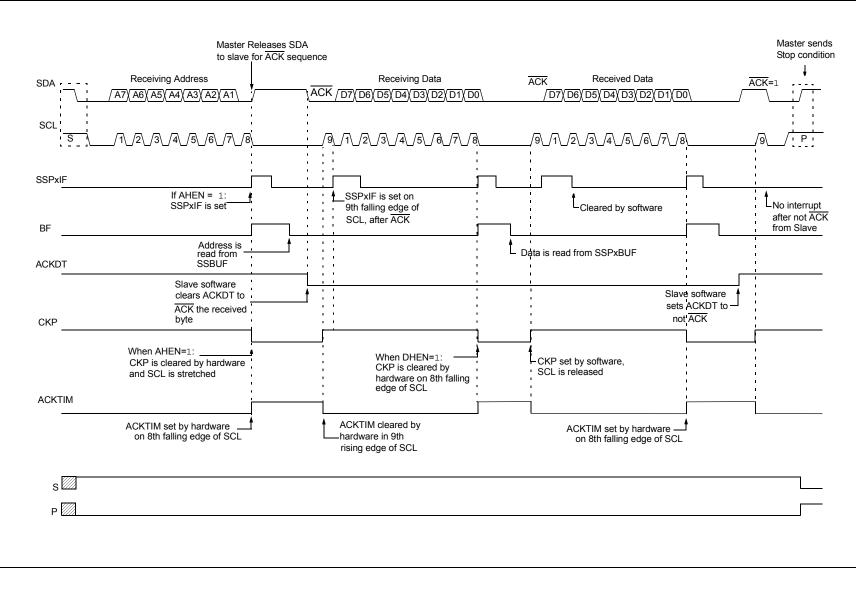


FIGURE 32-15: I²C SLAVE, 7-BIT ADDRESS, RECEPTION (SEN = 1, AHEN = 0, DHEN = 0)





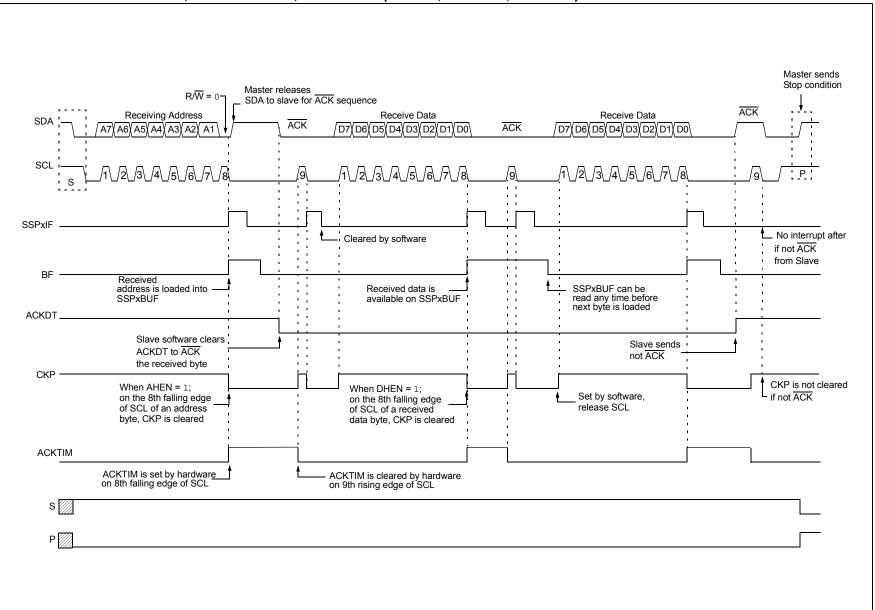


FIGURE 32-17: I²C SLAVE, 7-BIT ADDRESS, RECEPTION (SEN = 1, AHEN = 1, DHEN = 1)

32.5.3 SLAVE TRANSMISSION

When the R/W bit of the incoming address byte is set and an address match occurs, the R/W bit of the SSPxSTAT register is set. The received address is loaded into the SSPxBUF register, and an ACK pulse is sent by the slave on the ninth bit.

Following the ACK, slave hardware clears the CKP bit and the SCL pin is held low (see Section 32.5.6 "Clock Stretching" for more detail). By stretching the clock, the master will be unable to assert another clock pulse until the slave is done preparing the transmit data.

The transmit data must be loaded into the SSPxBUF register which also loads the SSPSR register. Then the SCL pin should be released by setting the CKP bit of the SSPxCON1 register. The eight data bits are shifted out on the falling edge of the SCL input. This ensures that the SDA signal is valid during the SCL high time.

The ACK pulse from the master-receiver is latched on the rising edge of the ninth SCL input pulse. This ACK value is copied to the ACKSTAT bit of the SSPxCON2 register. If ACKSTAT is set (not ACK), then the data transfer is complete. In this case, when the not ACK is latched by the slave, the slave goes idle and waits for another occurrence of the Start bit. If the SDA line was low (ACK), the next transmit data must be loaded into the SSPxBUF register. Again, the SCL pin must be released by setting bit CKP.

An MSSP interrupt is generated for each data transfer byte. The SSPxIF bit must be cleared by software and the SSPxSTAT register is used to determine the status of the byte. The SSPxIF bit is set on the falling edge of the ninth clock pulse.

32.5.3.1 Slave Mode Bus Collision

A slave receives a Read request and begins shifting data out on the SDA line. If a bus collision is detected and the SBCDE bit of the SSPxCON3 register is set, the BCLIF bit of the PIR register is set. Once a bus collision is detected, the slave goes idle and waits to be addressed again. User software can use the BCLIF bit to handle a slave bus collision.

32.5.3.2 7-bit Transmission

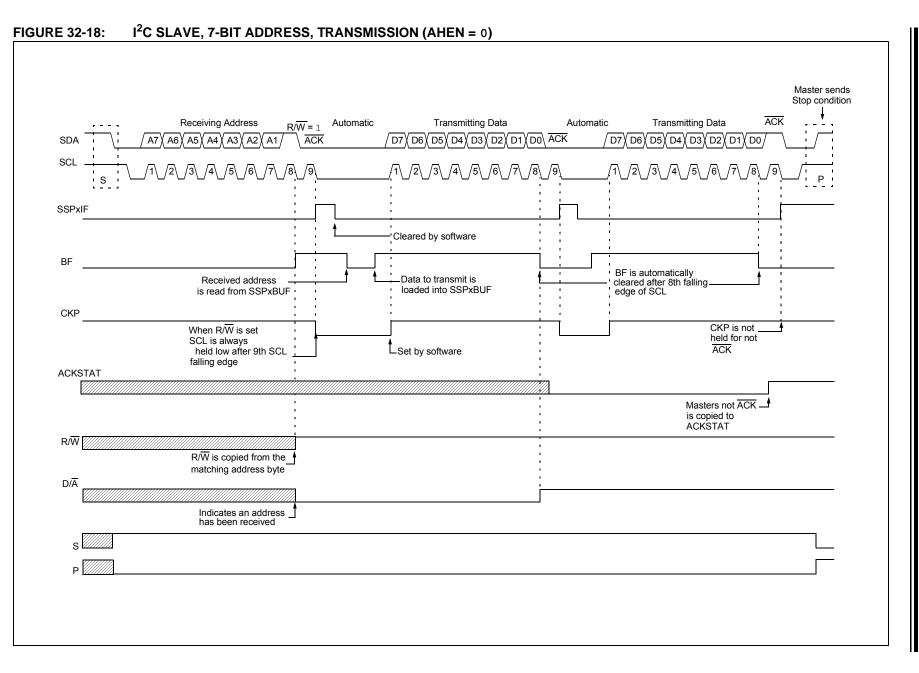
A master device can transmit a read request to a slave, and then clock data out of the slave. The list below outlines what software for a slave will need to do to accomplish a standard transmission. Figure 32-18 can be used as a reference to this list.

- 1. Master sends a Start condition on SDA and SCL.
- 2. S bit of SSPxSTAT is set; SSPxIF is set if interrupt on Start detect is enabled.
- Matching address with R/W bit set is received by the Slave setting SSPxIF bit.
- 4. Slave hardware generates an ACK and sets SSPxIF.
- 5. SSPxIF bit is cleared by user.
- 6. Software reads the received address from SSPxBUF, clearing BF.
- 7. R/\overline{W} is set so CKP was automatically cleared after the ACK.
- 8. The slave software loads the transmit data into SSPxBUF.
- 9. CKP bit is set releasing SCL, allowing the master to clock the data out of the slave.
- 10. SSPxIF is set after the ACK response from the master is loaded into the ACKSTAT register.
- 11. SSPxIF bit is cleared.
- 12. The slave software checks the ACKSTAT bit to see if the master wants to clock out more data.

Note 1: If the master ACKs the clock will be stretched.

2: ACKSTAT is the only bit updated on the rising edge of SCL (9th) rather than the falling.

- 13. Steps 9-13 are repeated for each transmitted byte.
- 14. If the master sends a not ACK; the clock is not held, but SSPxIF is still set.
- 15. The master sends a Restart condition or a Stop.
- 16. The slave is no longer addressed.



32.5.3.3 7-bit Transmission with Address Hold Enabled

Setting the AHEN bit of the SSPxCON3 register enables additional clock stretching and interrupt generation after the eighth falling edge of a received matching address. Once a matching address has been clocked in, CKP is cleared and the SSPxIF interrupt is set.

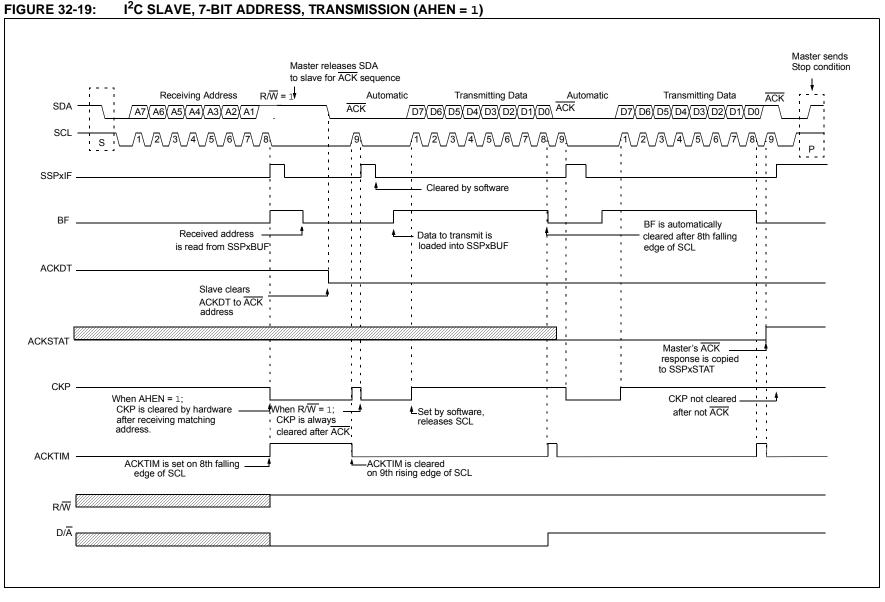
Figure 32-19 displays a standard waveform of a 7-bit address slave transmission with AHEN enabled.

- 1. Bus starts Idle.
- Master sends Start condition; the S bit of SSPxSTAT is set; SSPxIF is set if interrupt on Start detect is enabled.
- Master sends matching address with R/W bit set. After the eighth falling edge of the SCL line the CKP bit is cleared and SSPxIF interrupt is generated.
- 4. Slave software clears SSPxIF.
- Slave software reads ACKTIM bit of the SSPxCON3 register, and R/W and D/A of the SSPxSTAT register to determine the source of the interrupt.
- 6. Slave reads the address value from the SSPxBUF register clearing the BF bit.
- 7. Slave software decides from this information if it wishes to ACK or not ACK and sets the ACKDT bit of the SSPxCON2 register accordingly.
- 8. Slave sets the CKP bit releasing SCL.
- 9. Master clocks in the \overline{ACK} value from the slave.
- 10. Slave hardware automatically clears the CKP bit and sets SSPxIF after the ACK if the R/W bit is set.
- 11. Slave software clears SSPxIF.
- 12. Slave loads value to transmit to the master into SSPxBUF setting the BF bit.

Note: <u>SSPxBUF</u> cannot be loaded until after the ACK.

- 13. Slave sets the CKP bit releasing the clock.
- 14. Master clocks out the data from the slave and sends an ACK value on the ninth SCL pulse.
- 15. Slave hardware copies the ACK value into the ACKSTAT bit of the SSPxCON2 register.
- 16. Steps 10-15 are repeated for each byte transmitted to the master from the slave.
- 17. If the master sends a not ACK the slave releases the bus allowing the master to send a Stop and end the communication.

Note: Master must send a not ACK on the last byte to ensure that the slave releases the SCL line to receive a Stop.



32.5.4 SLAVE MODE 10-BIT ADDRESS RECEPTION

This section describes a standard sequence of events for the MSSP module configured as an I^2C slave in 10-bit Addressing mode.

Figure 32-20 is used as a visual reference for this description.

This is a step by step process of what must be done by slave software to accomplish I²C communication.

- 1. Bus starts Idle.
- Master sends Start condition; S bit of SSPxSTAT is set; SSPxIF is set if Interrupt on Start detect is enabled.
- Master sends matching high address with R/W bit clear; UA bit of the SSPxSTAT register is set.
- 4. Slave sends ACK and SSPxIF is set.
- 5. Software clears the SSPxIF bit.
- 6. Software reads received address from SSPxBUF clearing the BF flag.
- 7. Slave loads low address into SSPxADD, releasing SCL.
- 8. Master sends matching low address byte to the slave; UA bit is set.

Note: Updates to the SSPxADD register are not allowed until after the ACK sequence.

9. Slave sends ACK and SSPxIF is set.

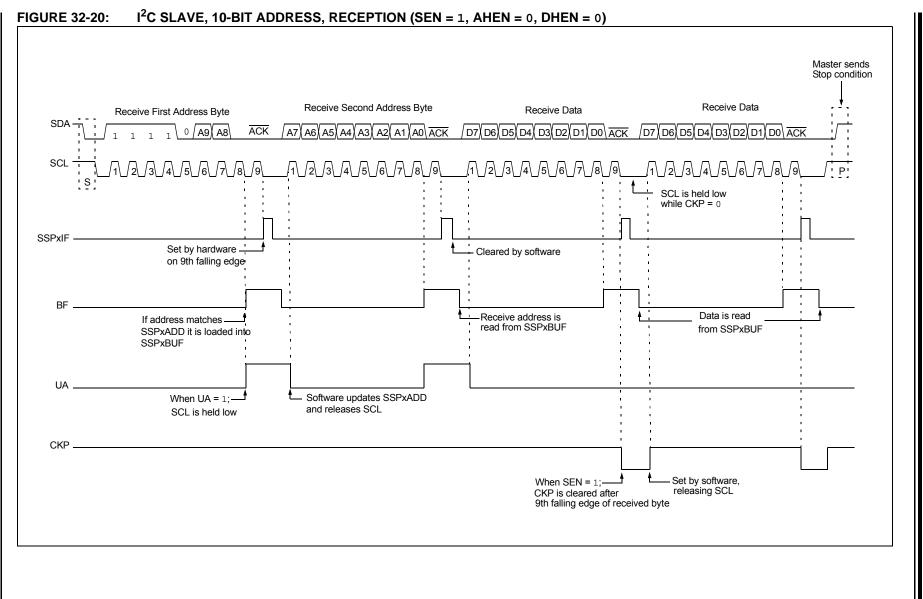
Note: If the low address does not match, SSPxIF and UA are still set so that the slave software can set SSPxADD back to the high address. BF is not set because there is no match. CKP is unaffected.

- 10. Slave clears SSPxIF.
- 11. Slave reads the received matching address from SSPxBUF clearing BF.
- 12. Slave loads high address into SSPxADD.
- 13. Master clocks a data byte to the slave and clocks out the slaves ACK on the ninth SCL pulse; SSPxIF is set.
- 14. If SEN bit of SSPxCON2 is set, CKP is cleared by hardware and the clock is stretched.
- 15. Slave clears SSPxIF.
- 16. Slave reads the received byte from SSPxBUF clearing BF.
- 17. If SEN is set the slave sets CKP to release the SCL.
- 18. Steps 13-17 repeat for each received byte.
- 19. Master sends Stop to end the transmission.

32.5.5 10-BIT ADDRESSING WITH ADDRESS OR DATA HOLD

Reception using 10-bit addressing with AHEN or DHEN set is the same as with 7-bit modes. The only difference is the need to update the SSPxADD register using the UA bit. All functionality, specifically when the CKP bit is cleared and SCL line is held low are the same. Figure 32-21 can be used as a reference of a slave in 10-bit addressing with AHEN set.

Figure 32-22 shows a standard waveform for a slave transmitter in 10-bit Addressing mode.



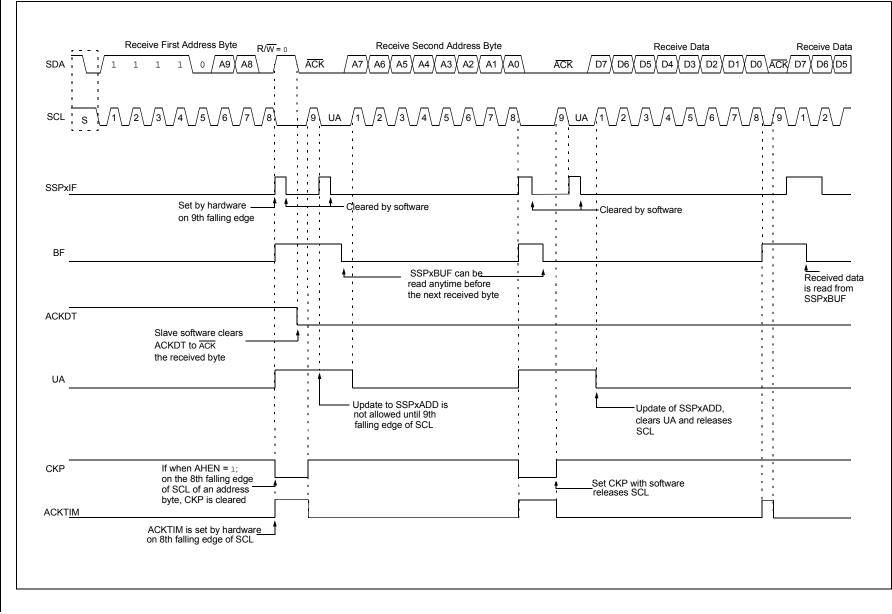


FIGURE 32-21: I²C SLAVE, 10-BIT ADDRESS, RECEPTION (SEN = 0, AHEN = 1, DHEN = 0)

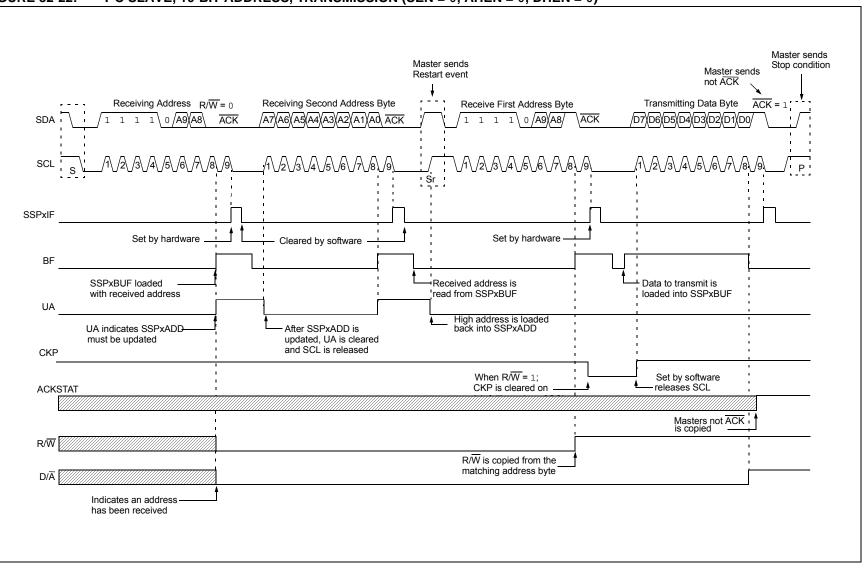


FIGURE 32-22: I^2C SLAVE, 10-BIT ADDRESS, TRANSMISSION (SEN = 0, AHEN = 0, DHEN = 0)

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32.5.6 CLOCK STRETCHING

Clock stretching occurs when a device on the bus holds the SCL line low, effectively pausing communication. The slave may stretch the clock to allow more time to handle data or prepare a response for the master device. A master device is not concerned with stretching as anytime it is active on the bus and not transferring data it is stretching. Any stretching done by a slave is invisible to the master software and handled by the hardware that generates SCL.

The CKP bit of the SSPxCON1 register is used to control stretching in software. Any time the CKP bit is cleared, the module will wait for the SCL line to go low and then hold it. Setting CKP will release SCL and allow more communication.

32.5.6.1 Normal Clock Stretching

Following an ACK if the R/W bit of SSPxSTAT is set, a read request, the slave hardware will clear CKP. This allows the slave time to update SSPxBUF with data to transfer to the master. If the SEN bit of SSPxCON2 is set, the slave hardware will always stretch the clock after the ACK sequence. Once the slave is ready; CKP is set by software and communication resumes.

- Note 1: The BF bit has no effect on if the clock will be stretched or not. This is different than previous versions of the module that would not stretch the clock, clear CKP, if SSPxBUF was read before the ninth falling edge of SCL.
 - 2: Previous versions of the module did not stretch the clock for a transmission if SSPxBUF was loaded before the ninth falling edge of SCL. It is now always cleared for read requests.

32.5.6.2 10-bit Addressing Mode

In 10-bit Addressing mode, when the UA bit is set the clock is always stretched. This is the only time the SCL is stretched without CKP being cleared. SCL is released immediately after a write to SSPxADD.

Note:	Previous versions of the module did not
	stretch the clock if the second address byte
	did not match.

32.5.6.3 Byte NACKing

When AHEN bit of SSPxCON3 is set; CKP is cleared by hardware after the eighth falling edge of SCL for a received matching address byte. When DHEN bit of SSPxCON3 is set; CKP is cleared after the eighth falling edge of SCL for received data.

Stretching after the eighth falling edge of SCL allows the slave to look at the received address or data and decide if it wants to ACK the received data.

32.5.6.4 Clock Synchronization and the CKP Bit

Any time the CKP bit is cleared, the module will wait for the SCL line to go low and then hold it. However, clearing the CKP bit will not assert the SCL output low until the SCL output is already sampled low. Therefore, the CKP bit will not assert the SCL line until an external I^2C master device has already asserted the SCL line. The SCL output will remain low until the CKP bit is set and all other devices on the I^2C bus have released SCL. This ensures that a write to the CKP bit will not violate the minimum high time requirement for SCL (see Figure 32-23).

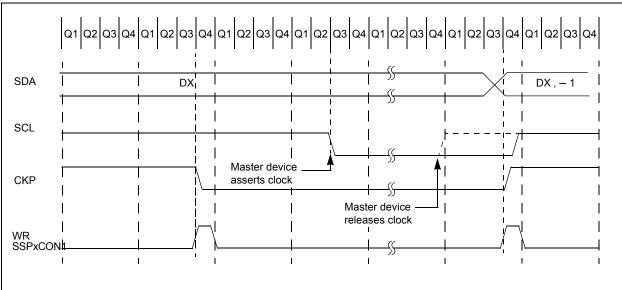


FIGURE 32-23: CLOCK SYNCHRONIZATION TIMING

32.5.7 GENERAL CALL ADDRESS SUPPORT

The addressing procedure for the I^2C bus is such that the first byte after the Start condition usually determines which device will be the slave addressed by the master device. The exception is the general call address which can address all devices. When this address is used, all devices should, in theory, respond with an acknowledge.

The general call address is a reserved address in the I²C protocol, defined as address 0x00. When the GCEN bit of the SSPxCON2 register is set, the slave module will automatically ACK the reception of this address regardless of the value stored in SSPxADD. After the slave clocks in an address of all zeros with the R/W bit clear, an interrupt is generated and slave software can read SSPxBUF and respond. Figure 32-24 shows reception а general call sequence.

In 10-bit Address mode, the UA bit will not be set on the reception of the general call address. The slave will prepare to receive the second byte as data, just as it would in 7-bit mode. If the AHEN bit of the SSPxCON3 register is set, just as with any other address reception, the slave hardware will stretch the clock after the eighth falling edge of SCL. The slave must then set its ACKDT value and release the clock with communication progressing as it would normally.

32.5.8 SSP MASK REGISTER

An SSP Mask (SSPxMSK) register (Register 32-5) is available in I²C Slave mode as a mask for the value held in the SSPSR register during an address comparison operation. A zero ('0') bit in the SSPxMSK register has the effect of making the corresponding bit of the received address a "don't care".

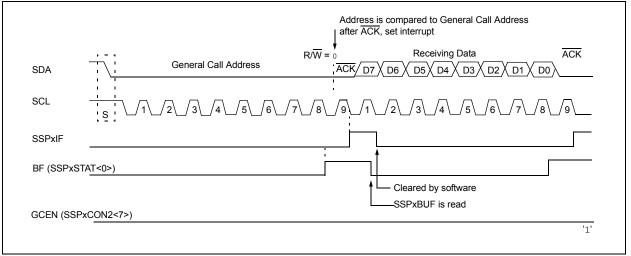
This register is reset to all '1's upon any Reset condition and, therefore, has no effect on standard SSP operation until written with a mask value.

The SSP Mask register is active during:

• 7-bit Address mode: address compare of A<7:1>.

10-bit Address mode: address compare of A<7:0> only. The SSP mask has no effect during the reception of the first (high) byte of the address.





32.6 I²C Master Mode

Master mode is enabled by setting and clearing the appropriate SSPM bits in the SSPxCON1 register and by setting the SSPEN bit. In Master mode, the SDA and SCK pins must be configured as inputs. The MSSP peripheral hardware will override the output driver TRIS controls when necessary to drive the pins low.

Master mode of operation is supported by interrupt generation on the detection of the Start and Stop conditions. The Stop (P) and Start (S) bits are cleared from a Reset or when the MSSP module is disabled. Control of the I^2C bus may be taken when the P bit is set, or the bus is Idle.

In Firmware Controlled Master mode, user code conducts all I²C bus operations based on Start and Stop bit condition detection. Start and Stop condition detection is the only active circuitry in this mode. All other communication is done by the user software directly manipulating the SDA and SCL lines.

The following events will cause the SSP Interrupt Flag bit, SSPxIF, to be set (SSP interrupt, if enabled):

- Start condition detected
- · Stop condition detected
- · Data transfer byte transmitted/received
- Acknowledge transmitted/received
- Repeated Start generated
 - Note 1: The MSSP module, when configured in I²C Master mode, does not allow queuing of events. For instance, the user is not allowed to initiate a Start condition and immediately write the SSPxBUF register to initiate transmission before the Start condition is complete. In this case, the SSPxBUF will not be written to and the WCOL bit will be set, indicating that a write to the SSPxBUF did not occur
 - 2: Master mode suspends Start/Stop detection when sending the Start/Stop condition by means of the SEN/PEN control bits. The SSPxIF bit is set at the end of the Start/Stop generation when hardware clears the control bit.

32.6.1 I²C MASTER MODE OPERATION

The master device generates all of the serial clock pulses and the Start and Stop conditions. A transfer is ended with a Stop condition or with a Repeated Start condition. Since the Repeated Start condition is also the beginning of the next serial transfer, the I²C bus will not be released.

In Master Transmitter mode, serial data is output through SDA, while SCL outputs the serial clock. The first byte transmitted contains the slave address of the receiving device (7 bits) and the Read/Write (R/W) bit. In this case, the R/W bit will be logic '0'. Serial data is transmitted eight bits at a time. After each byte is transmitted, an Acknowledge bit is received. Start and Stop conditions are output to indicate the beginning and the end of a serial transfer.

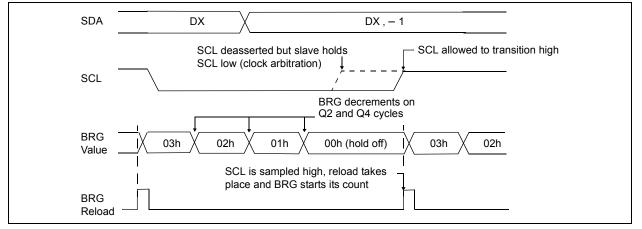
In Master Receive mode, the first byte transmitted contains the slave address of the transmitting device (7 bits) and the R/W bit. In this case, the R/W bit will be logic '1'. Thus, the first byte transmitted is a 7-bit slave address followed by a '1' to indicate the receive bit. Serial data is received via SDA, while SCL outputs the serial clock. Serial data is received eight bits at a time. After each byte is received, an Acknowledge bit is transmitted. Start and Stop conditions indicate the beginning and end of transmission.

A Baud Rate Generator is used to set the clock frequency output on SCL. See Section 32.7 "Baud Rate Generator" for more detail.

32.6.2 CLOCK ARBITRATION

Clock arbitration occurs when the master, during any receive, transmit or Repeated Start/Stop condition, releases the SCL pin (SCL allowed to float high). When the SCL pin is allowed to float high, the Baud Rate Generator (BRG) is suspended from counting until the SCL pin is actually sampled high. When the SCL pin is sampled high, the Baud Rate Generator is reloaded with the contents of SSPxADD<7:0> and begins counting. This ensures that the SCL high time will always be at least one BRG rollover count in the event that the clock is held low by an external device (Figure 32-25).

FIGURE 32-25: BAUD RATE GENERATOR TIMING WITH CLOCK ARBITRATION



32.6.3 WCOL STATUS FLAG

If the user writes the SSPxBUF when a Start, Restart, Stop, Receive or Transmit sequence is in progress, the WCOL bit is set and the contents of the buffer are unchanged (the write does not occur). Any time the WCOL bit is set it indicates that an action on SSPxBUF was attempted while the module was not idle.

Note:	Because queuing of events is not allowed,				
	writing to the lower five bits of SSPxCON2				
	is disabled until the Start condition is complete.				

32.6.4 I²C MASTER MODE START CONDITION TIMING

To initiate a Start condition (Figure 32-26), the user sets the Start Enable bit, SEN bit of the SSPxCON2 register. If the SDA and SCL pins are sampled high, the Baud Rate Generator is reloaded with the contents of SSPxADD<7:0> and starts its count. If SCL and SDA are both sampled high when the Baud Rate Generator times out (TBRG), the SDA pin is driven low. The action of the SDA being driven low while SCL is high is the Start condition and causes the S bit of the SSPxSTAT1 register to be set. Following this, the Baud Rate Generator is reloaded with the contents of SSPxADD<7:0> and resumes its count. When the Baud Rate Generator times out (TBRG), the SEN bit of the SSPxCON2 register will be automatically cleared

by hardware; the Baud Rate Generator is suspended, leaving the SDA line held low and the Start condition is complete.

- Note 1: If at the beginning of the Start condition, the SDA and SCL pins are already sampled low, or if during the Start condition, the SCL line is sampled low before the SDA line is driven low, a bus collision occurs, the Bus Collision Interrupt Flag, BCLIF, is set, the Start condition is aborted and the I²C module is reset into its Idle state.
 - 2: The Philips I²C specification states that a bus collision cannot occur on a Start.

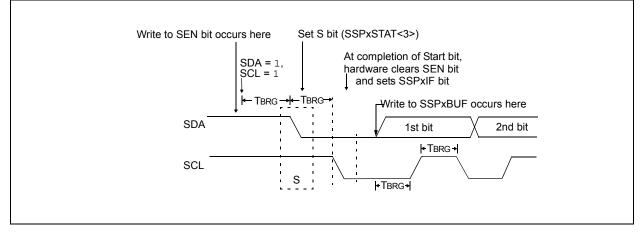


FIGURE 32-26: FIRST START BIT TIMING

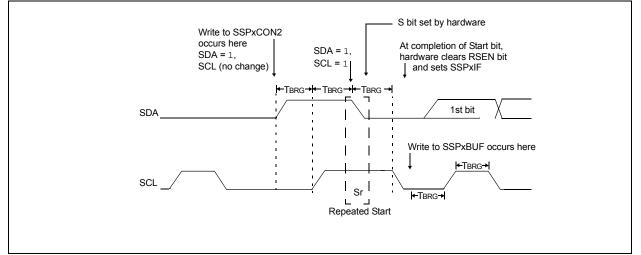
32.6.5 I²C MASTER MODE REPEATED START CONDITION TIMING

A Repeated Start condition (Figure 32-27) occurs when the RSEN bit of the SSPxCON2 register is programmed high and the master state machine is no longer active. When the RSEN bit is set, the SCL pin is asserted low. When the SCL pin is sampled low, the Baud Rate Generator is loaded and begins counting. The SDA pin is released (brought high) for one Baud Rate Generator count (TBRG). When the Baud Rate Generator times out, if SDA is sampled high, the SCL pin will be deasserted (brought high). When SCL is sampled high, the Baud Rate Generator is reloaded and begins counting. SDA and SCL must be sampled high for one TBRG. This action is then followed by assertion of the SDA pin (SDA = 0) for one TBRG while SCL is high. SCL is asserted low. Following this, the RSEN bit of the SSPxCON2 register will be automatically cleared and the Baud Rate Generator will not be reloaded, leaving the SDA pin held low. As soon as a Start condition is detected on the SDA and SCL pins, the S bit of the SSPxSTAT register will be set. The SSPxIF bit will not be set until the Baud Rate Generator has timed out.

Note 1: If RSEN is programmed while any other event is in progress, it will not take effect.

- **2:** A bus collision during the Repeated Start condition occurs if:
 - SDA is sampled low when SCL goes from low-to-high.
 - SCL goes low before SDA is asserted low. This may indicate that another master is attempting to transmit a data '1'.

FIGURE 32-27: REPEATED START CONDITION WAVEFORM



32.6.6 I²C MASTER MODE TRANSMISSION

Transmission of a data byte, a 7-bit address or the other half of a 10-bit address is accomplished by simply writing a value to the SSPxBUF register. This action will set the Buffer Full flag bit, BF, and allow the Baud Rate Generator to begin counting and start the next transmission. Each bit of address/data will be shifted out onto the SDA pin after the falling edge of SCL is asserted. SCL is held low for one Baud Rate Generator rollover count (TBRG). Data should be valid before SCL is released high. When the SCL pin is released high, it is held that way for TBRG. The data on the SDA pin must remain stable for that duration and some hold time after the next falling edge of SCL. After the eighth bit is shifted out (the falling edge of the eighth clock), the BF flag is cleared and the master releases SDA. This allows the slave device being addressed to respond with an \overline{ACK} bit during the ninth bit time if an address match occurred, or if data was received properly. The status of ACK is written into the ACKSTAT bit on the rising edge of the ninth clock. If the master receives an Acknowledge, the Acknowledge Status bit, ACKSTAT, is cleared. If not, the bit is set. After the ninth clock, the SSPxIF bit is set and the master clock (Baud Rate Generator) is suspended until the next data byte is loaded into the SSPxBUF, leaving SCL low and SDA unchanged (Figure 32-28).

After the write to the SSPxBUF, each bit of the address will be shifted out on the falling edge of SCL until all seven address bits and the R/W bit are completed. On the falling edge of the eighth clock, the master will release the SDA pin, allowing the slave to respond with an Acknowledge. On the falling edge of the ninth clock, the master will sample the SDA pin to see if the address was recognized by a slave. The status of the ACK bit is loaded into the ACKSTAT Status bit of the SSPxCON2 register. Following the falling edge of the ninth clock transmission of the address, the SSPxIF is set, the BF flag is cleared and the Baud Rate Generator is turned off until another write to the SSPxBUF takes place, holding SCL low and allowing SDA to float.

32.6.6.1 BF Status Flag

In Transmit mode, the BF bit of the SSPxSTAT register is set when the CPU writes to SSPxBUF and is cleared when all eight bits are shifted out.

32.6.6.2 WCOL Status Flag

If the user writes the SSPxBUF when a transmit is already in progress (i.e., SSPSR is still shifting out a data byte), the WCOL bit is set and the contents of the buffer are unchanged (the write does not occur).

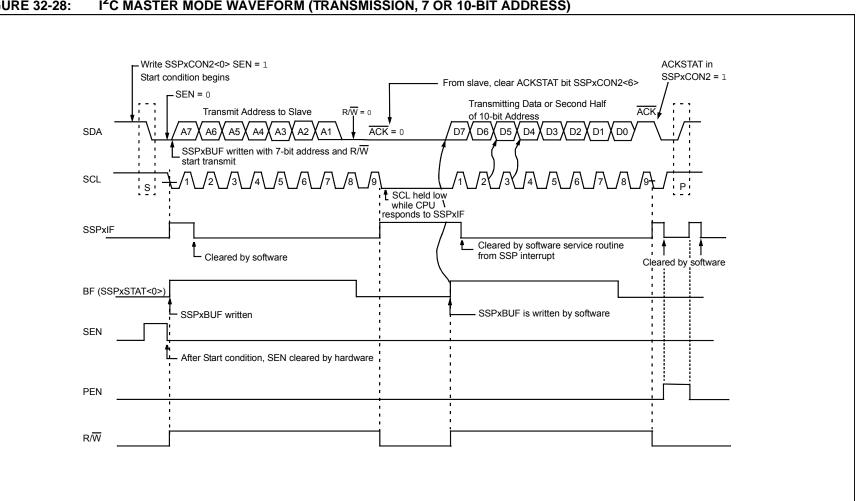
WCOL must be cleared by software before the next transmission.

32.6.6.3 ACKSTAT Status Flag

In Transmit mode, the ACKSTAT bit of the SSPxCON2 register is cleared when the slave has sent an Acknowledge ($\overrightarrow{ACK} = 0$) and is set when the slave does not Acknowledge ($\overrightarrow{ACK} = 1$). A slave sends an Acknowledge when it has recognized its address (including a general call), or when the slave has properly received its data.

32.6.6.4 Typical Transmit Sequence:

- 1. The user generates a Start condition by setting the SEN bit of the SSPxCON2 register.
- 2. SSPxIF is set by hardware on completion of the Start.
- 3. SSPxIF is cleared by software.
- 4. The MSSP module will wait the required start time before any other operation takes place.
- 5. The user loads the SSPxBUF with the slave address to transmit.
- 6. Address is shifted out the SDA pin until all eight bits are transmitted. Transmission begins as soon as SSPxBUF is written to.
- The MSSP module shifts in the ACK bit from the slave device and writes its value into the ACKSTAT bit of the SSPxCON2 register.
- The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSPxIF bit.
- 9. The user loads the SSPxBUF with eight bits of data.
- 10. Data is shifted out the SDA pin until all eight bits are transmitted.
- 11. The MSSP module shifts in the ACK bit from the slave device and writes its value into the ACKSTAT bit of the SSPxCON2 register.
- 12. Steps 8-11 are repeated for all transmitted data bytes.
- 13. The user generates a Stop or Restart condition by setting the PEN or RSEN bits of the SSPxCON2 register. Interrupt is generated once the Stop/Restart condition is complete.



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32.6.7 I²C MASTER MODE RECEPTION

Master mode reception (Figure 32-29) is enabled by programming the Receive Enable bit, RCEN bit of the SSPxCON2 register.

Note:	The MSSP module must be in an Idle
	state before the RCEN bit is set or the
	RCEN bit will be disregarded.

The Baud Rate Generator begins counting and on each rollover, the state of the SCL pin changes (high-to-low/low-to-high) and data is shifted into the SSPSR. After the falling edge of the eighth clock, the receive enable flag is automatically cleared, the contents of the SSPSR are loaded into the SSPxBUF, the BF flag bit is set, the SSPxIF flag bit is set and the Baud Rate Generator is suspended from counting, holding SCL low. The MSSP is now in Idle state awaiting the next command. When the buffer is read by the CPU, the BF flag bit is automatically cleared. The user can then send an Acknowledge bit at the end of reception by setting the Acknowledge Sequence Enable, ACKEN bit of the SSPxCON2 register.

32.6.7.1 BF Status Flag

In receive operation, the BF bit is set when an address or data byte is loaded into SSPxBUF from SSPSR. It is cleared when the SSPxBUF register is read.

32.6.7.2 SSPOV Status Flag

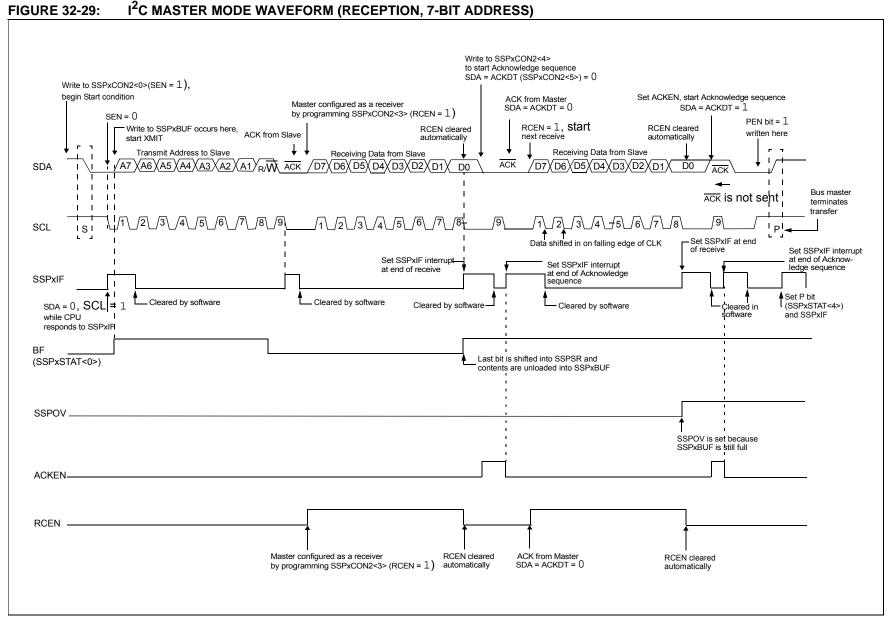
In receive operation, the SSPOV bit is set when eight bits are received into the SSPSR and the BF flag bit is already set from a previous reception.

32.6.7.3 WCOL Status Flag

If the user writes the SSPxBUF when a receive is already in progress (i.e., SSPSR is still shifting in a data byte), the WCOL bit is set and the contents of the buffer are unchanged (the write does not occur).

32.6.7.4 Typical Receive Sequence:

- 1. The user generates a Start condition by setting the SEN bit of the SSPxCON2 register.
- 2. SSPxIF is set by hardware on completion of the Start.
- 3. SSPxIF is cleared by software.
- 4. User writes SSPxBUF with the slave address to transmit and the R/W bit set.
- 5. Address is shifted out the SDA pin until all eight bits are transmitted. Transmission begins as soon as SSPxBUF is written to.
- 6. The MSSP module shifts in the ACK bit from the slave device and writes its value into the ACKSTAT bit of the SSPxCON2 register.
- 7. The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSPxIF bit.
- 8. User sets the RCEN bit of the SSPxCON2 register and the master clocks in a byte from the slave.
- 9. After the eighth falling edge of SCL, SSPxIF and BF are set.
- 10. Master clears SSPxIF and reads the received byte from SSPxBUF, clears BF.
- 11. Master sets ACK value sent to slave in ACKDT bit of the SSPxCON2 register and initiates the ACK by setting the ACKEN bit.
- 12. Master's ACK is clocked out to the slave and SSPxIF is set.
- 13. User clears SSPxIF.
- 14. Steps 8-13 are repeated for each received byte from the slave.
- 15. Master sends a not ACK or Stop to end communication.



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32.6.8 ACKNOWLEDGE SEQUENCE TIMING

An Acknowledge sequence is enabled by setting the Acknowledge Sequence Enable bit, ACKEN bit of the SSPxCON2 register. When this bit is set, the SCL pin is pulled low and the contents of the Acknowledge data bit are presented on the SDA pin. If the user wishes to generate an Acknowledge, then the ACKDT bit should be cleared. If not, the user should set the ACKDT bit before starting an Acknowledge sequence. The Baud Rate Generator then counts for one rollover period (TBRG) and the SCL pin is deasserted (pulled high). When the SCL pin is sampled high (clock arbitration), the Baud Rate Generator counts for TBRG. The SCL pin is then pulled low. Following this, the ACKEN bit is automatically cleared, the Baud Rate Generator is turned off and the MSSP module then goes into Idle mode (Figure 32-30).

32.6.8.1 WCOL Status Flag

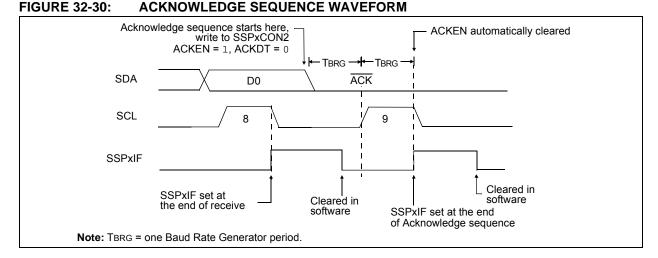
If the user writes the SSPxBUF when an Acknowledge sequence is in progress, then the WCOL bit is set and the contents of the buffer are unchanged (the write does not occur).

32.6.9 STOP CONDITION TIMING

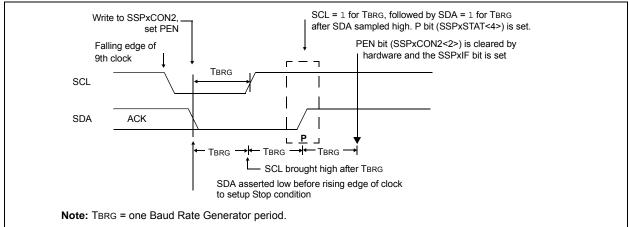
A Stop bit is asserted on the SDA pin at the end of a receive/transmit by setting the Stop Sequence Enable bit, PEN bit of the SSPxCON2 register. At the end of a receive/transmit, the SCL line is held low after the falling edge of the ninth clock. When the PEN bit is set, the master will assert the SDA line low. When the SDA line is sampled low, the Baud Rate Generator is reloaded and counts down to '0'. When the Baud Rate Generator times out, the SCL pin will be brought high and one TBRG (Baud Rate Generator rollover count) later, the SDA pin will be deasserted. When the SDA pin is sampled high while SCL is high, the P bit of the SSPxSTAT register is set. A TBRG later, the PEN bit is cleared and the SSPxIF bit is set (Figure 32-31).

32.6.9.1 WCOL Status Flag

If the user writes the SSPxBUF when a Stop sequence is in progress, then the WCOL bit is set and the contents of the buffer are unchanged (the write does not occur).







32.6.10 SLEEP OPERATION

While in Sleep mode, the I²C slave module can receive addresses or data and when an address match or complete byte transfer occurs, wake the processor from Sleep (if the MSSP interrupt is enabled).

32.6.11 EFFECTS OF A RESET

A Reset disables the MSSP module and terminates the current transfer.

32.6.12 MULTI-MASTER MODE

In Multi-Master mode, the interrupt generation on the detection of the Start and Stop conditions allows the determination of when the bus is free. The Stop (P) and Start (S) bits are cleared from a Reset or when the MSSP module is disabled. Control of the I²C bus may be taken when the P bit of the SSPxSTAT register is set, or the bus is Idle, with both the S and P bits clear. When the bus is busy, enabling the SSP interrupt will generate the interrupt when the Stop condition occurs.

In multi-master operation, the SDA line must be monitored for arbitration to see if the signal level is the expected output level. This check is performed by hardware with the result placed in the BCLIF bit.

The states where arbitration can be lost are:

- · Address Transfer
- Data Transfer
- A Start Condition
- A Repeated Start Condition
- An Acknowledge Condition

32.6.13 MULTI -MASTER COMMUNICATION, BUS COLLISION AND BUS ARBITRATION

Multi-Master mode support is achieved by bus arbitration. When the master outputs address/data bits onto the SDA pin, arbitration takes place when the master outputs a '1' on SDA, by letting SDA float high and another master asserts a '0'. When the SCL pin floats high, data should be stable. If the expected data on SDA is a '1' and the data sampled on the SDA pin is '0', then a bus collision has taken place. The master will set the Bus Collision Interrupt Flag, BCLIF and reset the I²C port to its Idle state (Figure 32-32).

If a transmit was in progress when the bus collision occurred, the transmission is halted, the BF flag is cleared, the SDA and SCL lines are deasserted and the SSPxBUF can be written to. When the user services the bus collision Interrupt Service Routine and if the I²C bus is free, the user can resume communication by asserting a Start condition.

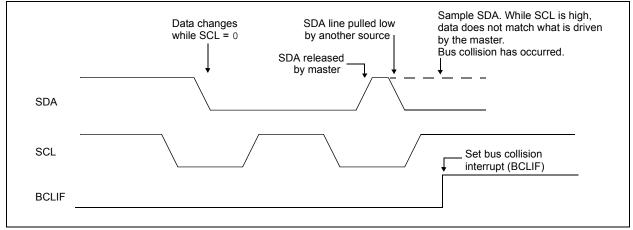
If a Start, Repeated Start, Stop or Acknowledge condition was in progress when the bus collision occurred, the condition is aborted, the SDA and SCL lines are deasserted and the respective control bits in the SSPxCON2 register are cleared. When the user services the bus collision Interrupt Service Routine and if the I²C bus is free, the user can resume communication by asserting a Start condition.

The master will continue to monitor the SDA and SCL pins. If a Stop condition occurs, the SSPxIF bit will be set.

A write to the SSPxBUF will start the transmission of data at the first data bit, regardless of where the transmitter left off when the bus collision occurred.

In Multi-Master mode, the interrupt generation on the detection of Start and Stop conditions allows the determination of when the bus is free. Control of the I^2C bus can be taken when the P bit is set in the SSPxSTAT register, or the bus is Idle and the S and P bits are cleared.

FIGURE 32-32: BUS COLLISION TIMING FOR TRANSMIT AND ACKNOWLEDGE



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32.6.13.1 Bus Collision During a Start Condition

During a Start condition, a bus collision occurs if:

- a) SDA or SCL are sampled low at the beginning of the Start condition (Figure 32-33).
- b) SCL is sampled low before SDA is asserted low (Figure 32-34).

During a Start condition, both the SDA and the SCL pins are monitored.

If the SDA pin is already low, or the SCL pin is already low, then all of the following occur:

- the Start condition is aborted,
- the BCLIF flag is set and
- the MSSP module is reset to its Idle state (Figure 32-33).

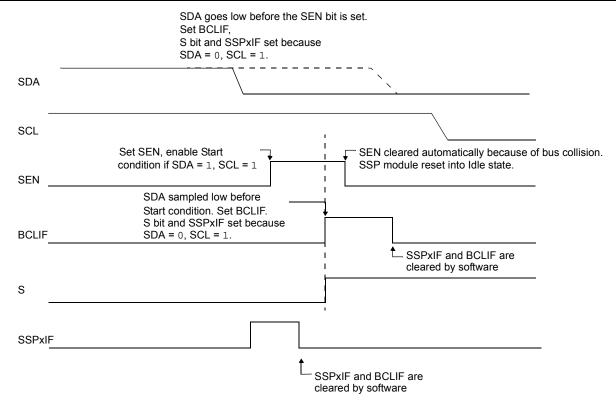
The Start condition begins with the SDA and SCL pins deasserted. When the SDA pin is sampled high, the Baud Rate Generator is loaded and counts down. If the SCL pin is sampled low while SDA is high, a bus

collision occurs because it is assumed that another master is attempting to drive a data '1' during the Start condition.

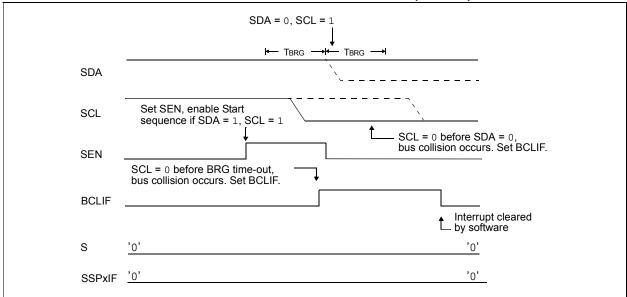
If the SDA pin is sampled low during this count, the BRG is reset and the SDA line is asserted early (Figure 32-35). If, however, a '1' is sampled on the SDA pin, the SDA pin is asserted low at the end of the BRG count. The Baud Rate Generator is then reloaded and counts down to zero; if the SCL pin is sampled as '0' during this time, a bus collision does not occur. At the end of the BRG count, the SCL pin is asserted low.

Note: The reason that bus collision is not a factor during a Start condition is that no two bus masters can assert a Start condition at the exact same time. Therefore, one master will always assert SDA before the other. This condition does not cause a bus collision because the two masters must be allowed to arbitrate the first address following the Start condition. If the address is the same, arbitration must be allowed to continue into the data portion, Repeated Start or Stop conditions.

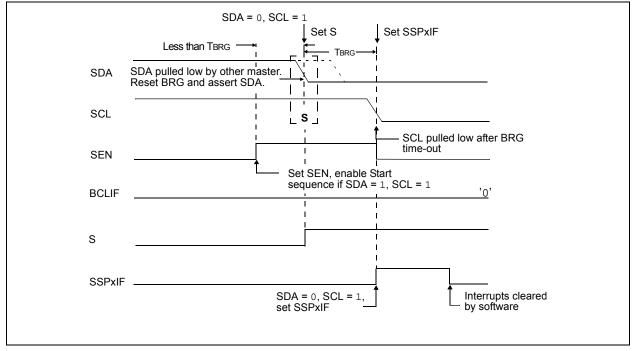












32.6.13.2 Bus Collision During a Repeated Start Condition

During a Repeated Start condition, a bus collision occurs if:

- a) A low level is sampled on SDA when SCL goes from low level to high level (Case 1).
- b) SCL goes low before SDA is asserted low, indicating that another master is attempting to transmit a data '1' (Case 2).

When the user releases SDA and the pin is allowed to float high, the BRG is loaded with SSPxADD and counts down to zero. The SCL pin is then deasserted and when sampled high, the SDA pin is sampled. If SDA is low, a bus collision has occurred (i.e., another master is attempting to transmit a data '0', Figure 32-36). If SDA is sampled high, the BRG is reloaded and begins counting. If SDA goes from high-to-low before the BRG times out, no bus collision occurs because no two masters can assert SDA at exactly the same time.

If SCL goes from high-to-low before the BRG times out and SDA has not already been asserted, a bus collision occurs. In this case, another master is attempting to transmit a data '1' during the Repeated Start condition, see Figure 32-37.

If, at the end of the BRG time-out, both SCL and SDA are still high, the SDA pin is driven low and the BRG is reloaded and begins counting. At the end of the count, regardless of the status of the SCL pin, the SCL pin is driven low and the Repeated Start condition is complete.

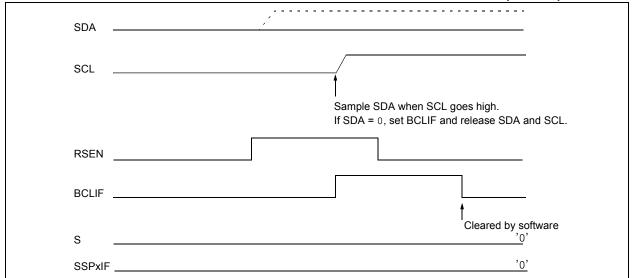
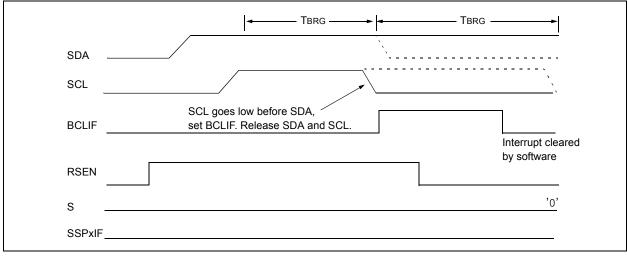


FIGURE 32-36: BUS COLLISION DURING A REPEATED START CONDITION (CASE 1)





32.6.13.3 Bus Collision During a Stop Condition

Bus collision occurs during a Stop condition if:

- a) After the SDA pin has been deasserted and allowed to float high, SDA is sampled low after the BRG has timed out (Case 1).
- b) After the SCL pin is deasserted, SCL is sampled low before SDA goes high (Case 2).

The Stop condition begins with SDA asserted low. When SDA is sampled low, the SCL pin is allowed to float. When the pin is sampled high (clock arbitration), the Baud Rate Generator is loaded with SSPxADD and counts down to zero. After the BRG times out, SDA is sampled. If SDA is sampled low, a bus collision has occurred. This is due to another master attempting to drive a data '0' (Figure 32-38). If the SCL pin is sampled low before SDA is allowed to float high, a bus collision occurs. This is another case of another master attempting to drive a data '0' (Figure 32-39).

FIGURE 32-38: BUS COLLISION DURING A STOP CONDITION (CASE 1)

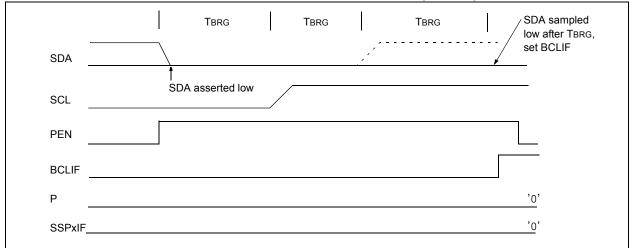
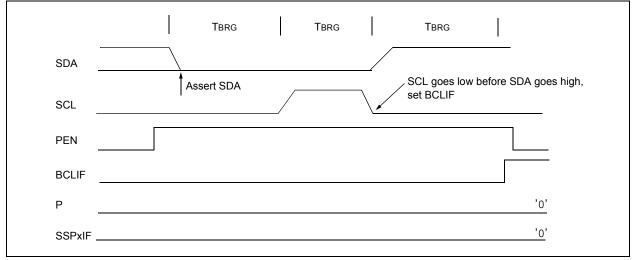


FIGURE 32-39: BUS COLLISION DURING A STOP CONDITION (CASE 2)



PIC16(L)F1777/8/9

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page:
ANSELA	—	—	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0	177
ANSELB	—	—	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	182
ANSELC	ANSC7	ANSC6	ANSC5	ANSC4	ANSC3	ANSC2	—		187
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	132
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	CCP1IE	TMR2IE	TMR1IE	133
PIE2	OSFIE	C2IE	C1IE	COG1IE	BCL1IE	C4IE	C3IE	CCP2IE	134
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	CCP1IF	TMR2IF	TMR1IF	139
PIR2	OSFIF	C2IF	C1IF	COG1IF	BCL1IF	C4IF	C3IF	CCP2IF	140
RxyPPS	—	—	RxyPPS<5:0>						
SSPCLKPPS	—	—			SSPCLK	PPS<5:0>			205, 207
SSPDATPPS	—	—			SSPDAT	PPS<5:0>			205, 207
SSPSSPPS	_	—			SSPSSF	PS<5:0>			205, 207
SSP1ADD				ADD	<7:0>				492
SSP1BUF	Synchronous	s Serial Port F	Receive Buffe	r/Transmit Re	egister				444*
SSP1CON1	WCOL	SSPOV	SSPEN	CKP		SSPN	l<3:0>		489
SSP1CON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	490
SSP1CON3	ACKTIM	PCIE	SCIE	BOEN	SDAHT	SBCDE	AHEN	DHEN	491
SSP1MSK				MSK	<7:0>				492
SSP1STAT	SMP	CKE	D/A	Р	S	R/W	UA	BF	488
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	176
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	181
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	186

TABLE 32-3: SUMMARY OF REGISTERS ASSOCIATED WITH I²C OPERATION

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by the MSSP module in I²C mode.

* Page provides register information.

32.7 BAUD RATE GENERATOR

The MSSP module has a Baud Rate Generator available for clock generation in both I²C and SPI Master modes. The Baud Rate Generator (BRG) reload value is placed in the SSPxADD register (Register 32-6). When a write occurs to SSPxBUF, the Baud Rate Generator will automatically begin counting down.

Once the given operation is complete, the internal clock will automatically stop counting and the clock pin will remain in its last state.

An internal signal "Reload" in Figure 32-40 triggers the value from SSPxADD to be loaded into the BRG counter. This occurs twice for each oscillation of the

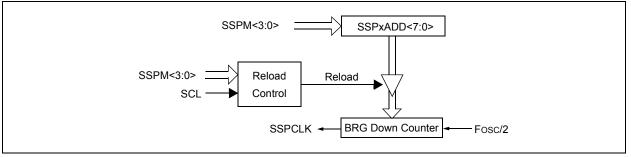
module clock line. The logic dictating when the reload signal is asserted depends on the mode the MSSP is being operated in.

Table 32-4demonstratesclockratesbasedoninstructioncyclesandtheBRGvalueloadedintoSSPxADD.

EQUATION 32-1:

$$FCLOCK = \frac{FOSC}{(SSPxADD + 1)(4)}$$

FIGURE 32-40: BAUD RATE GENERATOR BLOCK DIAGRAM



Note: Values of 0x00, 0x01 and 0x02 are not valid for SSPxADD when used as a Baud Rate Generator for I²C. This is an implementation limitation.

TABLE 32-4: MSSP CLOCK RATE W/BRG

Fosc	Fcy	BRG Value	FCLOCK (2 Rollovers of BRG)
32 MHz	8 MHz	13h	400 kHz
32 MHz	8 MHz	19h	308 kHz
32 MHz	8 MHz	4Fh	100 kHz
16 MHz	4 MHz	09h	400 kHz
16 MHz	4 MHz	0Ch	308 kHz
16 MHz	4 MHz	27h	100 kHz
4 MHz	1 MHz	09h	100 kHz

Note: Refer to the I/O port electrical specifications in Table 36-10 and Figure 32-7 to ensure the system is designed to support I/O requirements.

32.8 Register Definitions: MSSP Control

REGISTER 32-1: SSP1STAT: SSP STATUS REGISTER

R/W-0/0	R/W-0/0	R-0/0	R-0/0	R-0/0	R-0/0	R-0/0	R-0/0
SMP	CKE	D/A	Р	S	R/W	UA	BF
bit 7							
Legend:							
R = Readable bit		W = Writable bit		•	nted bit, read as '0		
u = Bit is unchan	iged	x = Bit is unknow		-n/n = Value at I	POR and BOR/Valu	ue at all other Res	sets
'1' = Bit is set		'0' = Bit is cleare	d				
bit 7	0 = Input data sa <u>SPI Slave mode</u> SMP must be cle <u>In I²C Master or</u> 1 = Slew rate co	e: ampled at end of d ampled at middle d 	of data output tin used in Slave n Standard Speed	node I mode (100 kHz	and 1 MHz)		
bit 6	 o = Slew rate control enabled for High-Speed mode (400 kHz) CKE: SPI Clock Edge Select bit (SPI mode only) In SPI Master or Slave mode: 1 = Transmit occurs on transition from active to Idle clock state o = Transmit occurs on transition from Idle to active clock state In I²C mode only: 1 = Enable input logic so that thresholds are compliant with SMBus specification o = Disable SMBus specific inputs 						
bit 5	1 = Indicates that	ess bit (I ² C mode o at the last byte rec at the last byte rec	eived or transmi				
bit 4	 0 = Indicates that the last byte received or transmitted was address P: Stop bit (I²C mode only. This bit is cleared when the MSSP module is disabled, SSPEN is cleared.) 1 = Indicates that a Stop bit has been detected last (this bit is '0' on Reset) 0 = Stop bit was not detected last 						
bit 3	1 = Indicates that	at a Start bit has be			ed, SSPEN is clea Reset)	red.)	
bit 2	 0 = Start bit was not detected last RW: Read/Write bit information (l²C mode only) This bit holds the R/W bit information following the last address match. This bit is only valid from the address match to the next Start bit, Stop bit, or not ACK bit. In l²C Slave mode: 1 = Read 0 = Write In l²C Master mode: 1 = Transmit is in progress 0 = Transmit is not in progress OR-ing this bit with SEN, RSEN, PEN, RCEN or ACKEN will indicate if the MSSP is in Idle mode. 						
bit 1	UA: Update Address bit (10-bit I ² C mode only) 1 = Indicates that the user needs to update the address in the SSP1ADD register						
bit 0	• • •						

REGISTER 32-2: SSP1CON1: SSP CONTROL REGISTER 1

R/C/HS-0/	0 R/C/HS-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	
WCOL	SSPOV ⁽¹⁾	SSPEN	CKP	SSPM<3:0>				
oit 7								
.egend:								
R = Readable	bit	W = Writable bit		U = Unimplement	ed hit, read as '0'			
u = Bit is unch		x = Bit is unknow	'n			at all other Resets		
1' = Bit is set	0	'0' = Bit is cleare				C = User cleared		
			u	HS = Bit is set by	naruware	C = Oser cleared		
bit 7	Master mode: 1 = A write to to 0 = No collisiono Slave mode:	n BUF register is writter	·	d while the I ² C condi			to be started	
it 6	In SPI mode: 1 = A new byte Overflow c setting ove SSP1BUF 0 = No overflo In I ² C mode: 1 = A byte is r	an only occur in Slav rflow. In Master mod register (must be cle w eceived while the S cleared in software).	e SSP1BUF regist ve mode. In Slave e, the overflow bit eared in software).	er is still holding the pr mode, the user must is not set since each n r is still holding the p	read the SSP1BUF, ew reception (and tr	even if only transmi ransmission) is initiat	itting data, to avoided by writing to the	
it 5	In both modes, v In <u>SPI mode</u> : 1 = Enables se 0 = Disables s In I ² C mode: 1 = Enables the		ce of the serial port					
bit 4	CKP: Clock Pola In SPI mode: 1 = Idle state for 0 = Idle state for In I ² C Slave mou SCL release cor 1 = Enable clock 0 = Holds clock In I ² C Master m	 Disables serial port and configures these pins as I/O port pins CKP: Clock Polarity Select bit 						
bit 3-0	1111 = I ² C Slav 1110 = I ² C Slav 1101 = Reserve 1001 = Reserve 1011 = I ² C firmv 1010 = SPI Mas 1001 = Reserve 1000 = I ² C Mas 0111 = I ² C Slav 0110 = SPI Slav 0101 = SPI Slav 0011 = SPI Mas 0010 = SPI Mas 0001 = SPI Mas	e mode, 7-bit addre d ware controlled Mas ster mode, clock = F d ter mode, clock = F re mode, 10-bit addre e mode, 7-bit addre	ress with Start and ter mode (slave i losc/(4 * (SSP1A losc / (4 * (SSP1A losc / (4 * (SSP1A losc / (4 * (SSP1A losc / (4 * (SSP1A losc / (4 * (SSP1A losc / (5 pin co 2_match/2 losc / (6 losc / 16	d Stop bit interrupts o Stop bit interrupts er dle) DD+1)) ⁽⁵⁾ ADD+1)) ⁽⁴⁾ ntrol disabled, SS ca	nabled	in		
2:	In Master mode, the ov When enabled, these p to select the pins. When enabled, the SD	verflow bit is not set bins must be proper	since each new ly configured as i	nput or output. Use S	SSPSSPPS, SSPC	LKPPS, SSPDATP	PS, and RxyPPS	

- 4: SSP1ADD values of 0, 1 or 2 are not supported for I²C mode.
- 5: SSP1ADD value of '0' is not supported. Use SSPM = 0000 instead.

R/W-0/0	R-0/0	R/W-0/0	R/S/HS-0/0	R/S/HS-0/0	R/S/HS-0/0	R/S/HS-0/0	R/W/HS-0/0		
GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN		
bit 7	•					•	bit (
Legend:									
R = Readable	e bit	W = Writable	bit	U = Unimpler	mented bit, read	l as '0'			
u = Bit is uncl	hanged	x = Bit is unk	nown	-n/n = Value	at POR and BO	R/Value at all c	other Resets		
'1' = Bit is set		'0' = Bit is cle	eared	HC = Cleared	d by hardware	S = User set			
bit 7	GCEN: Gene	eral Call Enable	e bit (in I ² C Sla	ve mode only)					
		nterrupt when a call address di	•	ddress (0x00 d	or 00h) is receiv	ed in the SSPS	SR		
bit 6	ACKSTAT: A	cknowledge Si	atus bit (in I ² C	mode only)					
		edge was not r							
		edge was recei	_	1					
bit 5		•	a bit (in I ² C moo	de only)					
	In Receive m		user initiates a	n Acknowledg	je sequence at i	the end of a rea	noivo		
	1 = Not Ackn		user initiates a						
	0 = Acknowle	0							
bit 4	ACKEN: Ack	knowledge Seq	uence Enable I	bit (in I ² C Mas	ter mode only)				
	In Master Re								
		Acknowledge sequence on SDA and SCL pins, and transmit ACKDT data bit tically cleared by hardware.							
		ledge sequenc							
bit 3		÷ .	(in I ² C Master I	mode onlv)					
		Receive mode		, , , , , , , , , , , , , , , , , , ,					
	0 = Receive	idle							
bit 2	PEN: Stop C	ondition Enabl	e bit (in I ² C Ma	ster mode onl	y)				
		Release Contro	-						
	1 = Initiate S 0 = Stop con		n SDA and SC	L pins. Autom	atically cleared	by hardware.			
bit 1	RSEN: Repe	ated Start Con	dition Enable b	it (in I ² C Mast	er mode only)				
	1 = Initiate F	Repeated Start	condition on SI	DA and SCL p	ins. Automatica	lly cleared by h	nardware.		
	0 = Repeated Start condition Idle								
bit 0			e/Stretch Enab	le bit					
	In Master mo		n SDA and SC	l nine Autom	atically cleared	by bardware			
	0 = Start con		II SDA and SC	L pins. Autom		by hardware.			
	In Slave mod								
	1 = Clock str			ave transmit a	nd slave receive	e (stretch enabl	ed)		
Note 1: Fo	r bits ACKEN, I	RCEN, PEN, R	SEN, SEN: If th	ne l ² C module	is not in the IdI	e mode, this bi	t may not be		

REGISTER 32-3: SSP1CON2: SSP CONTROL REGISTER 2⁽¹⁾

Note 1: For bits ACKEN, RCEN, PEN, RSEN, SEN: If the I²C module is not in the Idle mode, this bit may not be set (no spooling) and the SSP1BUF may not be written (or writes to the SSP1BUF are disabled).

R-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	
ACKTIM ⁽³⁾	PCIE	SCIE	BOEN	SDAHT	SBCDE	AHEN	DHEN	
bit 7			•				bit (
Legend:	L:1		L :4			1 (0)		
R = Readable		W = Writable		•	mented bit, read			
u = Bit is unch	langed		x = Bit is unknown -n/n = Value at POR and BOR/Value at a					
'1' = Bit is set		'0' = Bit is cle	ared					
bit 7	ACKTIM: Ac	IM: Acknowledge Time Status bit (I ² C mode only) ⁽³⁾						
	1 = Indicates	the I ² C bus is	in an Acknowl	vledge sequence, set on eighth falling edge of SCL clock red on ninth rising edge of SCL clock				
bit 6		Condition Interru	-					
	1 = Enable ir	nterrupt on dete ection interrupts	ection of Stop of	ondition	,			
bit 5	SCIE: Start C	Condition Interri	upt Enable bit ((I ² C slave mod	de only)			
	1 = Enable ir	nterrupt on dete ection interrupt	ection of Start of	or Restart cond	• •			
bit 4		er Overwrite En						
	0 = If ne SSP In I ² C Master This bit i In I ² C Slave 1 = SSF state 0 = SSP	P1BUF updates w byte is received 1CON1 register <u>r mode and SP</u> s ignored. <u>mode:</u> P1BUF is update of the SSPOV P1BUF is only u	ived with BF b r is set, and th <u>I Master mode</u> ted and ACK is bit only if the pdated when S	it of the SSP1 e buffer is not <u>:</u> s generated fo BF bit = 0. SSPOV is clea	or a received ad	Iready set, SS	POV bit of th	
bit 3		A Hold Time Se		• •				
	0 = Minimum	n of 300 ns hold n of 100 ns hold	I time on SDA	after the falling	g edge of SCL			
bit 2	SBCDE: Slav	ve Mode Bus C	ollision Detect	Enable bit (I ²	C Slave mode c	only)		
		ing edge of SC f the PIR2 regis			en the module is	s outputting a l	nigh state, th	
		lave bus collision inter		bled				
bit 1	AHEN: Addr	ess Hold Enabl	e bit (l ² C Slave	e mode only)				
	 AHEN: Address Hold Enable bit (I²C Slave mode only) 1 = Following the eighth falling edge of SCL for a matching received address byte; CKP bit of the SSP1CON1 register will be cleared and the SCL will be held low. 0 = Address holding is disabled 							
bit 0	DHEN: Data	Hold Enable bi	it (I ² C Slave m	ode only)				
	bit of the	g the eighth fall SSP1CON1 ro ding is disabled	egister and SC		ed data byte; sla	ave hardware c	lears the CK	
who	⁻ daisy-chained en a new byte i	SPI operation; s received and	allows the use BF = 1, but ha	rdware continu	out the last recei les to write the n	nost recent byte	e to SSP1BUI	

REGISTER 32-4: SSP1CON3: SSP CONTROL REGISTER 3

2: This bit has no effect in Slave modes that Start and Stop condition detection is explicitly listed as enabled.3: The ACKTIM Status bit is only active when the AHEN bit or DHEN bit is set.

REGISTER 32-5: SSP1MSK: SSP MASK REGISTER

R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1		
			MSK	<7:0>					
bit 7							bit 0		
Legend:									
R = Readable bit		W = Writable	bit	U = Unimpler	nented bit, read	l as '0'			
u = Bit is unchanged		x = Bit is unkr	nown	-n/n = Value at POR and BOR/Value at all other Resets					
'1' = Bit is set '0' = Bit is cleared		ared							
bit 7-1	MSK<7:1>:								
	 1 = The received address bit n is compared to SSP1ADD<n> to detect I²C address match</n> 0 = The received address bit n is not used to detect I²C address match 								
bit 0	_	ask bit for I ² C S	•						
	I ² C Slave mode, 10-bit address (SSPM<3:0> = 0111 or 1111): 1 = The received address bit 0 is compared to SSP1ADD<0> to detect I ² C address match						. (.).		
						r-C address m	atch		
	0 = The received address bit 0 is not used to detect I^2C address match I^2C Slave mode. This address the bit is imported								

I²C Slave mode, 7-bit address, the bit is ignored

'0' = Bit is cleared

REGISTER 32-6: SSP1ADD: MSSP ADDRESS AND BAUD RATE REGISTER (I²C MODE)

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
			ADD	<7:0>			
bit 7							bit 0
Legend:							
R = Readable	= Readable bit W = Writable bit		pit	U = Unimpler	nented bit, read	d as '0'	
u = Bit is unch	anged	x = Bit is unkn	own	-n/n = Value a	at POR and BO	R/Value at all o	other Resets

Master mode:

1' = Bit is set

bit 7-0 ADD<7:0>: Baud Rate Clock Divider bits SCL pin clock period = ((ADD<7:0> + 1) *4)/Fosc

<u>10-Bit Slave mode – Most Significant Address Byte:</u>

- bit 7-3 **Not used:** Unused for Most Significant Address Byte. Bit state of this register is a "don't care". Bit pattern sent by master is fixed by I²C specification and must be equal to '11110'. However, those bits are compared by hardware and are not affected by the value in this register.
- bit 2-1 ADD<2:1>: Two Most Significant bits of 10-bit address
- bit 0 Not used: Unused in this mode. Bit state is a "don't care".

10-Bit Slave mode – Least Significant Address Byte:

bit 7-0 ADD<7:0>: Eight Least Significant bits of 10-bit address

7-Bit Slave mode:

	bit 7-1	ADD<7:1>: 7-bit address
--	---------	-------------------------

bit 0 Not used: Unused in this mode. Bit state is a "don't care".

33.0 ENHANCED UNIVERSAL SYNCHRONOUS ASYNCHRONOUS RECEIVER TRANSMITTER (EUSART)

The Enhanced Universal Synchronous Asynchronous Receiver Transmitter (EUSART) module is a serial I/O communications peripheral. It contains all the clock generators, shift registers and data buffers necessary to perform an input or output serial data transfer independent of device program execution. The EUSART, also known as a Serial Communications Interface (SCI), can be configured as a full-duplex asynchronous system or half-duplex synchronous system. Full-Duplex mode is useful for communications with peripheral systems, such as CRT terminals and personal computers. Half-Duplex Synchronous mode is intended for communications with peripheral devices, such as A/D or D/A integrated circuits, serial EEPROMs or other microcontrollers. These devices typically do not have internal clocks for baud rate generation and require the external clock signal provided by a master synchronous device.

The EUSART module includes the following capabilities:

- · Full-duplex asynchronous transmit and receive
- Two-character input buffer
- One-character output buffer
- · Programmable 8-bit or 9-bit character length
- · Address detection in 9-bit mode
- · Input buffer overrun error detection
- Received character framing error detection
- Half-duplex synchronous master
- · Half-duplex synchronous slave
- Programmable clock polarity in synchronous modes
- Sleep operation

The EUSART module implements the following additional features, making it ideally suited for use in Local Interconnect Network (LIN) bus systems:

- · Automatic detection and calibration of the baud rate
- Wake-up on Break reception
- · 13-bit Break character transmit

Block diagrams of the EUSART transmitter and receiver are shown in Figure 33-1 and Figure 33-2.

The EUSART transmit output (TX_out) is available to the TX/CK pin and internally to the following peripherals:

- Configurable Logic Cell (CLC)
- Data signal modulator (DSM)

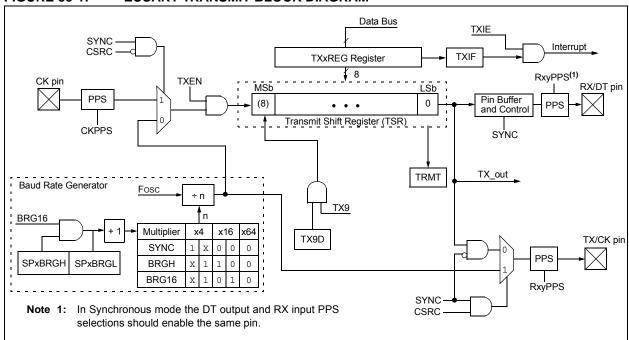
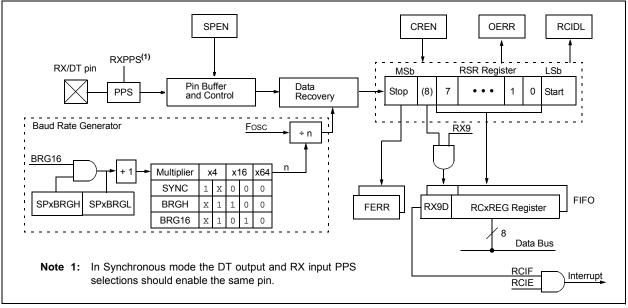


FIGURE 33-1: EUSART TRANSMIT BLOCK DIAGRAM





The operation of the EUSART module is controlled through three registers:

- Transmit Status and Control (TXxSTA)
- Receive Status and Control (RCxSTA)
- Baud Rate Control (BAUDxCON)

These registers are detailed in Register 33-1, Register 33-2 and Register 33-3, respectively.

The RX and CK input pins are selected with the RXPPS and CKPPS registers, respectively. TX, CK, and DT output pins are selected with each pin's RxyPPS register. Since the RX input is coupled with the DT output in Synchronous mode, it is the user's responsibility to select the same pin for both of these functions when operating in Synchronous mode. The EUSART control logic will control the data direction drivers automatically.

33.1 EUSART Asynchronous Mode

The EUSART transmits and receives data using the standard non-return-to-zero (NRZ) format. NRZ is implemented with two levels: a VOH Mark state which represents a '1' data bit, and a VoL Space state which represents a '0' data bit. NRZ refers to the fact that consecutively transmitted data bits of the same value stay at the output level of that bit without returning to a neutral level between each bit transmission. An NRZ transmission port idles in the Mark state. Each character transmission consists of one Start bit followed by eight or nine data bits and is always terminated by one or more Stop bits. The Start bit is always a space and the Stop bits are always marks. The most common data format is eight bits. Each transmitted bit persists for a period of 1/(baud rate). An on-chip dedicated 8-bit/16-bit Baud Rate Generator is used to derive standard baud rate frequencies from the system oscillator. See Table 33-5 for examples of baud rate configurations.

The EUSART transmits and receives the LSb first. The EUSART's transmitter and receiver are functionally independent, but share the same data format and baud rate. Parity is not supported by the hardware, but can be implemented in software and stored as the ninth data bit.

33.1.1 EUSART ASYNCHRONOUS TRANSMITTER

The EUSART transmitter block diagram is shown in Figure 33-1. The heart of the transmitter is the serial Transmit Shift Register (TSR), which is not directly accessible by software. The TSR obtains its data from the transmit buffer, which is the TXxREG register.

33.1.1.1 Enabling the Transmitter

The EUSART transmitter is enabled for asynchronous operations by configuring the following three control bits:

- TXEN = 1
- SYNC = 0
- SPEN = 1

All other EUSART control bits are assumed to be in their default state.

Setting the TXEN bit of the TXxSTA register enables the transmitter circuitry of the EUSART. Clearing the SYNC bit of the TXxSTA register configures the EUSART for asynchronous operation. Setting the SPEN bit of the RCxSTA register enables the EUSART and automatically configures the TX/CK I/O pin as an output. If the TX/CK pin is shared with an analog peripheral, the analog I/O function must be disabled by clearing the corresponding ANSEL bit.

Note: The TXIF Transmitter Interrupt flag is set when the TXEN enable bit is set.

33.1.1.2 Transmitting Data

A transmission is initiated by writing a character to the TXxREG register. If this is the first character, or the previous character has been completely flushed from the TSR, the data in the TXxREG is immediately transferred to the TSR register. If the TSR still contains all or part of a previous character, the new character data is held in the TXxREG until the Stop bit of the previous character has been transmitted. The pending character in the TXxREG is then transferred to the TSR in one TcY immediately following the Stop bit sequence commences immediately following the transfer of the data to the TSR from the TXxREG.

33.1.1.3 Transmit Data Polarity

The polarity of the transmit data can be controlled with the SCKP bit of the BAUDxCON register. The default state of this bit is '0', which selects high true transmit idle and data bits. Setting the SCKP bit to '1' will invert the transmit data resulting in low true idle and data bits. The SCKP bit controls transmit data polarity in Asynchronous mode only. In Synchronous mode, the SCKP bit has a different function. See Section 33.5.1.2 "Clock Polarity".

33.1.1.4 Transmit Interrupt Flag

The TXIF interrupt flag bit of the PIR1 register is set whenever the EUSART transmitter is enabled and no character is being held for transmission in the TXxREG. In other words, the TXIF bit is only clear when the TSR is busy with a character and a new character has been queued for transmission in the TXxREG. The TXIF flag bit is not cleared immediately upon writing TXxREG. TXIF becomes valid in the second instruction cycle following the write execution. Polling TXIF immediately following the TXxREG write will return invalid results. The TXIF bit is read-only, it cannot be set or cleared by software.

The TXIF interrupt can be enabled by setting the TXIE interrupt enable bit of the PIE1 register. However, the TXIF flag bit will be set whenever the TXxREG is empty, regardless of the state of the TXIE enable bit.

To use interrupts when transmitting data, set the TXIE bit only when there is more data to send. Clear the TXIE interrupt enable bit upon writing the last character of the transmission to the TXxREG.

33.1.1.5 TSR Status

The TRMT bit of the TXxSTA register indicates the status of the TSR register. This is a read-only bit. The TRMT bit is set when the TSR register is empty and is cleared when a character is transferred to the TSR register from the TXxREG. The TRMT bit remains clear until all bits have been shifted out of the TSR register. No interrupt logic is tied to this bit, so the user has to poll this bit to determine the TSR status.

Note:	The TSR register is not mapped in data
	memory, so it is not available to the user.

33.1.1.6 Transmitting 9-Bit Characters

The EUSART supports 9-bit character transmissions. When the TX9 bit of the TXxSTA register is set, the EUSART will shift nine bits out for each character transmitted. The TX9D bit of the TXxSTA register is the ninth, and Most Significant data bit. When transmitting 9-bit data, the TX9D data bit must be written before writing the eight Least Significant bits into the TXxREG. All nine bits of data will be transferred to the TSR shift register immediately after the TXxREG is written.

A special 9-bit Address mode is available for use with multiple receivers. See **Section 33.1.2.7** "Address **Detection**" for more information on the Address mode.

- 33.1.1.7 Asynchronous Transmission Set-up:
- Initialize the SPxBRGH:SPxBRGL register pair and the BRGH and BRG16 bits to achieve the desired baud rate (see Section 33.4 "EUSART Baud Rate Generator (BRG)").
- 2. Enable the asynchronous serial port by clearing the SYNC bit and setting the SPEN bit.
- 3. If 9-bit transmission is desired, set the TX9 control bit. A set ninth data bit will indicate that the eight Least Significant data bits are an address when the receiver is set for address detection.
- 4. Set the SCKP bit if inverted transmit is desired.
- 5. Enable the transmission by setting the TXEN control bit. This will cause the TXIF interrupt bit to be set.
- If interrupts are desired, set the TXIE interrupt enable bit of the PIE1 register. An interrupt will occur immediately provided that the GIE and PEIE bits of the INTCON register are also set.
- 7. If 9-bit transmission is selected, the ninth bit should be loaded into the TX9D data bit.
- 8. Load 8-bit data into the TXxREG register. This will start the transmission.

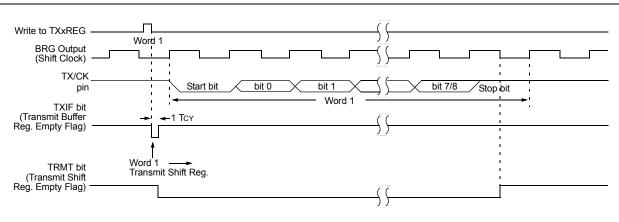


FIGURE 33-3: ASYNCHRONOUS TRANSMISSION



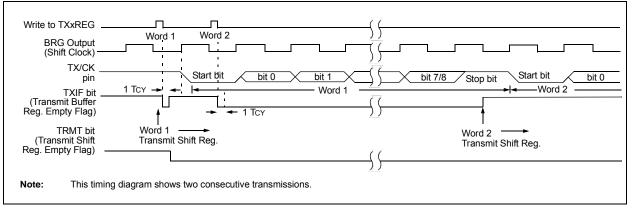


TABLE 33-1: SUMMARY OF REGISTERS ASSOCIATED WITH ASYNCHRONOUS TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ANSELA	—	—	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0	177
ANSELB	—	—	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	182
ANSELC	ANSC7	ANSC6	ANSC5	ANSC4	ANSC3	ANSC2		—	187
BAUD1CON	ABDOVF	RCIDL	_	SCKP	BRG16	_	WUE	ABDEN	505
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	132
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	CCP1IE	TMR2IE	TMR1IE	133
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	CCP1IF	TMR2IF	TMR1IF	139
RC1STA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	504
RxyPPS	_	RxyPPS<5:0>							
SP1BRGL	SP1BRG<7:0>								506*
SP1BRGH				SP1BRG	G<15:8>				506*
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	176
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	181
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	186
TX1REG	EUSART Transmit Data Register								495*
TX1STA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	503

Legend: — = unimplemented location, read as '0'. Shaded cells are not used for asynchronous transmission.

Page provides register information.

33.1.2 EUSART ASYNCHRONOUS RECEIVER

The Asynchronous mode is typically used in RS-232 systems. The receiver block diagram is shown in Figure 33-2. The data is received on the RX/DT pin and drives the data recovery block. The data recovery block is actually a high-speed shifter operating at 16 times the baud rate, whereas the serial Receive Shift Register (RSR) operates at the bit rate. When all eight or nine bits of the character have been shifted in, they are immediately transferred to a two character First-In-First-Out (FIFO) memory. The FIFO buffering allows reception of two complete characters and the start of a third character before software must start servicing the EUSART receiver. The FIFO and RSR registers are not directly accessible by software. Access to the received data is via the RCxREG register.

33.1.2.1 Enabling the Receiver

The EUSART receiver is enabled for asynchronous operation by configuring the following three control bits:

- CREN = 1
- SYNC = 0
- SPEN = 1

All other EUSART control bits are assumed to be in their default state.

Setting the CREN bit of the RCxSTA register enables the receiver circuitry of the EUSART. Clearing the SYNC bit of the TXxSTA register configures the EUSART for asynchronous operation. Setting the SPEN bit of the RCxSTA register enables the EUSART. The programmer must set the corresponding TRIS bit to configure the RX/DT I/O pin as an input.

Note: If the RX/DT function is on an analog pin, the corresponding ANSEL bit must be cleared for the receiver to function.

33.1.2.2 Receiving Data

The receiver data recovery circuit initiates character reception on the falling edge of the first bit. The first bit, also known as the Start bit, is always a zero. The data recovery circuit counts one-half bit time to the center of the Start bit and verifies that the bit is still a zero. If it is not a zero then the data recovery circuit aborts character reception, without generating an error, and resumes looking for the falling edge of the Start bit. If the Start bit zero verification succeeds then the data recovery circuit counts a full bit time to the center of the next bit. The bit is then sampled by a majority detect circuit and the resulting '0' or '1' is shifted into the RSR. This repeats until all data bits have been sampled and shifted into the RSR. One final bit time is measured and the level sampled. This is the Stop bit, which is always a '1'. If the data recovery circuit samples a '0' in the Stop bit position then a framing error is set for this character, otherwise the framing error is cleared for this character. See Section 33.1.2.4 "Receive Framing Error" for more information on framing errors.

Immediately after all data bits and the Stop bit have been received, the character in the RSR is transferred to the EUSART receive FIFO and the RCIF interrupt flag bit of the PIR1 register is set. The top character in the FIFO is transferred out of the FIFO by reading the RCxREG register.

Note:	If the receive FIFO is overrun, no additional									
	characters will be received until the overrun									
	condition is cleared. See Section 33.1.2.5									
	"Receive Overrun Error" for more									
	information on overrun errors.									

33.1.2.3 Receive Interrupts

The RCIF interrupt flag bit of the PIR1 register is set whenever the EUSART receiver is enabled and there is an unread character in the receive FIFO. The RCIF interrupt flag bit is read-only, it cannot be set or cleared by software.

RCIF interrupts are enabled by setting all of the following bits:

- RCIE, Interrupt Enable bit of the PIE1 register
- PEIE, Peripheral Interrupt Enable bit of the INTCON register
- GIE, Global Interrupt Enable bit of the INTCON register

The RCIF interrupt flag bit will be set when there is an unread character in the FIFO, regardless of the state of interrupt enable bits.

33.1.2.4 Receive Framing Error

Each character in the receive FIFO buffer has a corresponding framing error Status bit. A framing error indicates that a Stop bit was not seen at the expected time. The framing error status is accessed via the FERR bit of the RCxSTA register. The FERR bit represents the status of the top unread character in the receive FIFO. Therefore, the FERR bit must be read before reading the RCxREG.

The FERR bit is read-only and only applies to the top unread character in the receive FIFO. A framing error (FERR = 1) does not preclude reception of additional characters. It is not necessary to clear the FERR bit. Reading the next character from the FIFO buffer will advance the FIFO to the next character and the next corresponding framing error.

The FERR bit can be forced clear by clearing the SPEN bit of the RCxSTA register which resets the EUSART. Clearing the CREN bit of the RCxSTA register does not affect the FERR bit. A framing error by itself does not generate an interrupt.

Note:	If all receive characters in the receive									
	FIFO have framing errors, repeated reads									
	of the RCxREG will not clear the FERR bit.									

33.1.2.5 Receive Overrun Error

The receive FIFO buffer can hold two characters. An overrun error will be generated if a third character, in its entirety, is received before the FIFO is accessed. When this happens the OERR bit of the RCxSTA register is set. The characters already in the FIFO buffer can be read but no additional characters will be received until the error is cleared. The error must be cleared by either clearing the CREN bit of the RCxSTA register or by resetting the EUSART by clearing the SPEN bit of the RCxSTA register.

33.1.2.6 Receiving 9-Bit Characters

The EUSART supports 9-bit character reception. When the RX9 bit of the RCxSTA register is set the EUSART will shift nine bits into the RSR for each character received. The RX9D bit of the RCxSTA register is the ninth and Most Significant data bit of the top unread character in the receive FIFO. When reading 9-bit data from the receive FIFO buffer, the RX9D data bit must be read before reading the eight Least Significant bits from the RCxREG.

33.1.2.7 Address Detection

A special Address Detection mode is available for use when multiple receivers share the same transmission line, such as in RS-485 systems. Address detection is enabled by setting the ADDEN bit of the RCxSTA register.

Address detection requires 9-bit character reception. When address detection is enabled, only characters with the ninth data bit set will be transferred to the receive FIFO buffer, thereby setting the RCIF interrupt bit. All other characters will be ignored.

Upon receiving an address character, user software determines if the address matches its own. Upon address match, user software must disable address detection by clearing the ADDEN bit before the next Stop bit occurs. When user software detects the end of the message, determined by the message protocol used, software places the receiver back into the Address Detection mode by setting the ADDEN bit.

- 33.1.2.8 Asynchronous Reception Set-up
- Initialize the SPxBRGH:SPxBRGL register pair and the BRGH and BRG16 bits to achieve the desired baud rate (see Section 33.4 "EUSART Baud Rate Generator (BRG)").
- 2. Clear the ANSEL bit for the RX pin (if applicable).
- Enable the serial port by setting the SPEN bit. The SYNC bit must be clear for asynchronous operation.
- 4. If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
- 5. If 9-bit reception is desired, set the RX9 bit.
- 6. Enable reception by setting the CREN bit.
- 7. The RCIF interrupt flag bit will be set when a character is transferred from the RSR to the receive buffer. An interrupt will be generated if the RCIE interrupt enable bit was also set.
- 8. Read the RCxSTA register to get the error flags and, if 9-bit data reception is enabled, the ninth data bit.
- 9. Get the received eight Least Significant data bits from the receive buffer by reading the RCxREG register.
- 10. If an overrun occurred, clear the OERR flag by clearing the CREN receiver enable bit.

33.1.2.9 9-bit Address Detection Mode Set-up

This mode would typically be used in RS-485 systems. To set up an Asynchronous Reception with Address Detect Enable:

- Initialize the SPxBRGH:SPxBRGL register pair and the BRGH and BRG16 bits to achieve the desired baud rate (see Section 33.4 "EUSART Baud Rate Generator (BRG)").
- 2. Clear the ANSEL bit for the RX pin (if applicable).
- Enable the serial port by setting the SPEN bit. The SYNC bit must be clear for asynchronous operation.
- If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
- 5. Enable 9-bit reception by setting the RX9 bit.
- 6. Enable address detection by setting the ADDEN bit.
- 7. Enable reception by setting the CREN bit.
- The RCIF interrupt flag bit will be set when a character with the ninth bit set is transferred from the RSR to the receive buffer. An interrupt will be generated if the RCIE interrupt enable bit was also set.
- 9. Read the RCxSTA register to get the error flags. The ninth data bit will always be set.
- 10. Get the received eight Least Significant data bits from the receive buffer by reading the RCxREG register. Software determines if this is the device's address.
- 11. If an overrun occurred, clear the OERR flag by clearing the CREN receiver enable bit.
- 12. If the device has been addressed, clear the ADDEN bit to allow all received data into the receive buffer and generate interrupts.

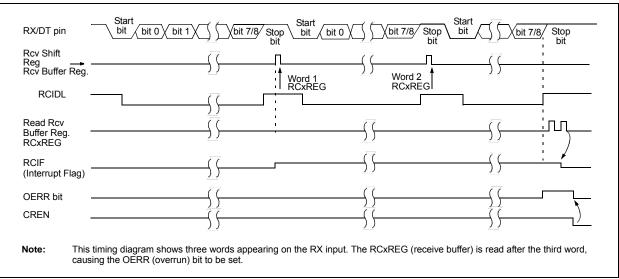


FIGURE 33-5: ASYNCHRONOUS RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ANSELA	—	—	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0	177
ANSELB	—	_	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	182
ANSELC	ANSC7	ANSC6	ANSC5	ANSC4	ANSC3	ANSC2	—	—	187
BAUD1CON	ABDOVF	RCIDL	_	SCKP	BRG16	—	WUE	ABDEN	505
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	132
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	CCP1IE	TMR2IE	TMR1IE	133
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	CCP1IF	TMR2IF	TMR1IF	139
RC1REG			EUS	SART Receiv	e Data Regis	ter			498*
RC1STA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	504
RxyPPS	—	— — RxyPPS<5:0>							205
SP1BRGL				SP1BR0	G<7:0>				506
SP1BRGH				SP1BRG	6<15:8>				506
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	176
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	181
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	186
TX1STA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	503

TABLE 33-2:	SUMMARY OF REGISTERS ASSOCIATED WITH ASYNCHRONOUS RECEPTION	
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Legend: — = unimplemented location, read as '0'. Shaded cells are not used for asynchronous reception.

* Page provides register information.

33.2 Clock Accuracy with Asynchronous Operation

The factory calibrates the internal oscillator block output (INTOSC). However, the INTOSC frequency may drift as VDD or temperature changes, and this directly affects the asynchronous baud rate. Two methods may be used to adjust the baud rate clock, but both require a reference clock source of some kind.

The first (preferred) method uses the OSCTUNE register to adjust the INTOSC output. Adjusting the value in the OSCTUNE register allows for fine resolution changes to the system clock source. See Section 5.2.2.3 "Internal Oscillator Frequency Adjustment" for more information.

The other method adjusts the value in the Baud Rate Generator. This can be done automatically with the Auto-Baud Detect feature (see Section 33.4.1 "Auto-Baud Detect"). There may not be fine enough resolution when adjusting the Baud Rate Generator to compensate for a gradual change in the peripheral clock frequency.

33.3 Register Definitions: EUSART Control

REGISTER 33-1: TX1STA: TRANSMIT STATUS AND CONTROL REGISTER

R/W-/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R-1/1	R/W-0/0				
CSRC	TX9	TXEN ⁽¹⁾	SYNC	SENDB	BRGH	TRMT	TX9D				
bit 7					•		bit (
Legend:	. 1.11		1.11			(0)					
R = Readable		W = Writable			mented bit, read		11. D. 1.				
u = Bit is unc	•	x = Bit is unki		-n/n = value a	at POR and BOF	R/Value at all c	other Resets				
'1' = Bit is se	L	'0' = Bit is cle	areo								
bit 7		< Source Select	bit								
	Asynchronou	<u>is mode</u> :									
	Don't care Synchronous	mode:									
	•	mode (clock ge	nerated intern	allv from BRG)						
		node (clock from			,						
bit 6	TX9: 9-bit Tr	ansmit Enable I	oit								
		9-bit transmiss									
		8-bit transmiss									
oit 5		TXEN: Transmit Enable bit ⁽¹⁾ 1 = Transmit enabled									
	0 = Transmi										
oit 4	SYNC: EUS	ART Mode Sele	ect bit								
		1 = Synchronous mode									
	0 = Asynchr										
bit 3		nd Break Chara	cter bit								
	<u>Asynchronous mode</u> : 1 = Send Sync Break on next transmission (cleared by hardware upon completion)										
		eak transmissio				ompiction					
	Synchronous		·								
	Don't care										
bit 2	-	Baud Rate Sel	ect bit								
	Asynchronou 1 = High spe										
	0 = Low spectrum										
		Synchronous mode:									
	Unused in th	is mode									
bit 1		mit Shift Regist	ter Status bit								
	1 = TSR em 0 = TSR full										
bit 0		bit of Transmit	Data								
		ess/data bit or a									
	San Se adar										

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R-0/0	R-0/0	R-0/0				
SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D				
bit 7					·		bit 0				
Legend:											
R = Readable	e bit	W = Writable	bit	U = Unimple	mented bit, read	l as '0'					
u = Bit is uncl	•	x = Bit is unkr	nown	-n/n = Value	at POR and BO	R/Value at all o	ther Resets				
'1' = Bit is set		'0' = Bit is cle	ared								
bit 7	SDEN: Sorial	Port Enable bi	ł								
	1 = Serial po		L								
		rt disabled (he	d in Reset)								
bit 6		eceive Enable I									
		9-bit reception 3-bit reception									
bit 5		e Receive Enat	ole bit								
	Asynchronou										
	Don't care										
	Synchronous mode – Master:										
	1 = Enables single receive										
	 Disables single receive This bit is cleared after reception is complete. 										
	<u>Synchronous mode – Slave</u>										
	Don't care										
bit 4	CREN: Conti	nuous Receive	Enable bit								
	<u>Asynchronou</u>	<u>s mode</u> :									
		1 = Enables receiver									
	0 = Disables receiver										
	<u>Synchronous mode</u> : 1 = Enables continuous receive until enable bit CREN is cleared (CREN overrides SREN)										
	 1 = Enables continuous receive until enable bit CREN is cleared (CREN overrides SREN) 0 = Disables continuous receive 										
bit 3	ADDEN: Add	ress Detect En	able bit								
	Asynchronous mode 9-bit (RX9 = 1):										
	1 = Enables address detection, enable interrupt and load the receive buffer when RSR<8> is set										
	0 = Disables address detection, all bytes are received and ninth bit can be used as parity bit										
	<u>Asynchronous mode 8-bit (RX9 = 0)</u> : Don't care										
hit 0		ng Error bit									
bit 2		FERR: Framing Error bit 1 = Framing error (can be updated by reading RCxREG register and receive next valid byte)									
	1 = Praining 0 = No frami				register and re		i byte)				
bit 1	OERR: Over	run Error bit									
		error (can be c	leared by clea	aring bit CREN)						
	0 = No overr		-								
bit 0		bit of Received				C.					
	This can be a	iddress/data bi	or a parity bi	t and must be	calculated by us	er firmware.					

REGISTER 33-2: RC1STA: RECEIVE STATUS AND CONTROL REGISTER

R-0/0	R-1/1	U-0	R/W-0/0	R/W-0/0	U-0	R/W-0/0	R/W-0/0
ABDOVF	RCIDL	_	SCKP	BRG16	—	WUE	ABDEN
bit 7							bit 0
Legend:							
R = Readable	e bit	W = Writable	bit	U = Unimple	mented bit, read	d as '0'	
u = Bit is uncl	hanged	x = Bit is unk	nown	-n/n = Value	at POR and BC	R/Value at all c	other Resets
'1' = Bit is set		'0' = Bit is cle	ared				
bit 7		uto-Baud Dete	t Overflow bit				
	Asynchronou						
	-	<u>id timer overflo</u>	wed				
		id timer did not					
	<u>Synchronous</u> Don't care	<u>s mode</u> :					
bit 6	RCIDL: Rece	eive Idle Flag b	it				
	<u>Asynchronou</u>	is mode:					
	1 = Receiver						
		nas been receiv	ed and the re	ceiver is recei	ving		
	Synchronous Don't care	<u>s mode</u> .					
bit 5	Unimplemer	nted: Read as	ʻ0'				
bit 4	SCKP: Syncl	hronous Clock	Polarity Select	t bit			
	Asynchronou	<u>is mode</u> :					
		inverted data t non-inverted d					
	<u>Synchronous</u>	<u>s mode</u> :					
		locked on rising locked on fallin					
bit 3	BRG16: 16-t	oit Baud Rate C	Generator bit				
		aud Rate Gene ud Rate Genera					
bit 2	Unimplemer	nted: Read as	0'				
bit 1	WUE: Wake-	up Enable bit					
	<u>Asynchronou</u>	<u>is mode</u> :					
		is waiting for a natically clear a			will be received	l, byte RCIF wil	l be set. WUE
		is operating n	ormally				
	Synchronous	<u>s mode</u> :					
	Don't care						
bit 0		o-Baud Detect	Enable bit				
	Asynchronou					nlata)	
		ud Detect mod ud Detect mod		liears when at	uto-baud is com	piete)	
	Synchronous		E IS UISADIEU				
	Don't care						

REGISTER 33-3: BAUD1CON: BAUD RATE CONTROL REGISTER

33.4 EUSART Baud Rate Generator (BRG)

The Baud Rate Generator (BRG) is an 8-bit or 16-bit timer that is dedicated to the support of both the asynchronous and synchronous EUSART operation. By default, the BRG operates in 8-bit mode. Setting the BRG16 bit of the BAUDxCON register selects 16-bit mode.

The SPxBRGH:SPxBRGL register pair determines the period of the free running baud rate timer. In Asynchronous mode the multiplier of the baud rate period is determined by both the BRGH bit of the TXxSTA register and the BRG16 bit of the BAUDxCON register. In Synchronous mode, the BRGH bit is ignored.

Table 33-3 contains the formulas for determining the baud rate. Example 33-1 provides a sample calculation for determining the baud rate and baud rate error.

Typical baud rates and error values for various Asynchronous modes have been computed for your convenience and are shown in Table 33-5. It may be advantageous to use the high baud rate (BRGH = 1), or the 16-bit BRG (BRG16 = 1) to reduce the baud rate error. The 16-bit BRG mode is used to achieve slow baud rates for fast oscillator frequencies.

Writing a new value to the SPxBRGH:SPxBRGL register pair causes the BRG timer to be reset (or cleared). This ensures that the BRG does not wait for a timer overflow before outputting the new baud rate.

If the system clock is changed during an active receive operation, a receive error or data loss may result. To avoid this problem, check the status of the RCIDL bit to make sure that the receive operation is idle before changing the system clock.

EXAMPLE 33-1: CALCULATING BAUD RATE ERROR

For a device with Fosc of 16 MHz, desired baud rate of 9600, Asynchronous mode, 8-bit BRG: Fosc Desired Baud Rate = $\frac{1000}{64([SPBRGH:SPBRGL] + 1)}$ Solving for SPxBRGH:SPxBRGL: Fosc $X = \frac{\overline{Desired Baud Rate}}{1} - 1$ 64 16000000 $\frac{9600}{64} - 1$ = [25.042] = 25 Calculated Baud Rate = $\frac{16000000}{64(25+1)}$ = 9615Error = Calc. Baud Rate – Desired Baud Rate Desired Baud Rate $= \frac{(9615 - 9600)}{9600} = 0.16\%$

(Configuration Bi	its		David Data Farmula
SYNC	BRG16	BRGH	BRG/EUSART Mode	Baud Rate Formula
0	0	0	8-bit/Asynchronous	Fosc/[64 (n+1)]
0	0	1	8-bit/Asynchronous	
0	1	0	16-bit/Asynchronous	Fosc/[16 (n+1)]
0	1	1	16-bit/Asynchronous	
1	0	x	8-bit/Synchronous	Fosc/[4 (n+1)]
1	1	x	16-bit/Synchronous	

TABLE 33-3: BAUD RATE FORMULAS

Legend: x = Don't care, n = value of SPxBRGH:SPxBRGL register pair.

TABLE 33-4: SUMMARY OF REGISTERS ASSOCIATED WITH THE BAUD RATE GENERATOR

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
BAUD1CON	ABDOVF	RCIDL	_	SCKP	BRG16	_	WUE	ABDEN	505
RC1STA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	504
SP1BRGL				SP1BR	G<7:0>				506
SP1BRGH				SP1BR0	G<15:8>				506
TX1STA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	503

Legend: — = unimplemented location, read as '0'. Shaded cells are not used for the Baud Rate Generator.

* Page provides register information.

					SYNC	C = 0, BRGH	l = 0, BRC	G16 = 0					
BAUD	Foso	; = 32.00	0 MHz	Fosc	Fosc = 20.000 MHz			: = 18.43	2 MHz	Fosc = 11.0592 MHz			
RATE	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)	
300	—	_	_	_	_	_	_	_	_	_	_	_	
1200	_	_	_	1221	1.73	255	1200	0.00	239	1200	0.00	143	
2400	2404	0.16	207	2404	0.16	129	2400	0.00	119	2400	0.00	71	
9600	9615	0.16	51	9470	-1.36	32	9600	0.00	29	9600	0.00	17	
10417	10417	0.00	47	10417	0.00	29	10286	-1.26	27	10165	-2.42	16	
19.2k	19.23k	0.16	25	19.53k	1.73	15	19.20k	0.00	14	19.20k	0.00	8	
57.6k	55.55k	-3.55	3	—	_	_	57.60k	0.00	7	57.60k	0.00	2	
115.2k	—	—	—	_	_	—	_	_	—	_	_	—	

TABLE 33-5: BAUD RATES FOR ASYNCHRONOUS MODES

					SYNC	C = 0, BRGH	l = 0, BRG	616 = 0				
BAUD	Fos	c = 8.00) MHz	Fos	c = 4.000) MHz	Fosc	= 3.686	4 MHz	Fos	c = 1.000) MHz
RATE	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)
300	_		_	300	0.16	207	300	0.00	191	300	0.16	51
1200	1202	0.16	103	1202	0.16	51	1200	0.00	47	1202	0.16	12
2400	2404	0.16	51	2404	0.16	25	2400	0.00	23	_	_	_
9600	9615	0.16	12	_	_	_	9600	0.00	5	_	_	_
10417	10417	0.00	11	10417	0.00	5	_	_	_	_	_	_
19.2k	—	_	_	_	_	_	19.20k	0.00	2	_	_	_
57.6k	—	—	—	_	_	—	57.60k	0.00	0	—	_	—
115.2k	—	—	_	—	_	_	_	_	—	—	_	—

					SYNC	C = 0, BRGH	l = 1, BRG	316 = 0					
BAUD	Foso	; = 32.00	0 MHz	Fosc	Fosc = 20.000 MHz			: = 18.43	2 MHz	Fosc = 11.0592 MHz			
RATE	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)	
300	—	—	_			—			_	_	—	—	
1200	—	—	—	—		—	—	—	—	—	—	—	
2400	_	_	_	_	_	_	_	_	_	—	_	_	
9600	9615	0.16	207	9615	0.16	129	9600	0.00	119	9600	0.00	71	
10417	10417	0.00	191	10417	0.00	119	10378	-0.37	110	10473	0.53	65	
19.2k	19.23k	0.16	103	19.23k	0.16	64	19.20k	0.00	59	19.20k	0.00	35	
57.6k	57.14k	-0.79	34	56.82k	-1.36	21	57.60k	0.00	19	57.60k	0.00	11	
115.2k	117.64k	2.12	16	113.64k	-1.36	10	115.2k	0.00	9	115.2k	0.00	5	

				-	SYNC	C = 0, BRGH	I = 1, BRG	616 = 0				
BAUD	Fos	c = 8.000) MHz	Fos	c = 4.000) MHz	Fosc	: = 3.686	4 MHz	Fos	c = 1.000) MHz
RATE	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)
300	_	_	_	_	_	_		_	_	300	0.16	207
1200	—	—	—	1202	0.16	207	1200	0.00	191	1202	0.16	51
2400	2404	0.16	207	2404	0.16	103	2400	0.00	95	2404	0.16	25
9600	9615	0.16	51	9615	0.16	25	9600	0.00	23	_	_	_
10417	10417	0.00	47	10417	0.00	23	10473	0.53	21	10417	0.00	5
19.2k	19231	0.16	25	19.23k	0.16	12	19.2k	0.00	11	_	_	_
57.6k	55556	-3.55	8	—	_	_	57.60k	0.00	3	—	_	_
115.2k	—	_	—	_	_	—	115.2k	0.00	1		_	—

TABLE 33-5: BAUD RATES FOR ASYNCHRONOUS MODES (CONTINUED)

					SYNC	C = 0, BRGH	l = 0, BRG	616 = 1					
BAUD	Foso	= 32.00	0 MHz	Fosc	Fosc = 20.000 MHz			: = 18.43	2 MHz	Fosc = 11.0592 MHz			
RATE	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)	
300	300.0	0.00	6666	300.0	-0.01	4166	300.0	0.00	3839	300.0	0.00	2303	
1200	1200	-0.02	3332	1200	-0.03	1041	1200	0.00	959	1200	0.00	575	
2400	2401	-0.04	832	2399	-0.03	520	2400	0.00	479	2400	0.00	287	
9600	9615	0.16	207	9615	0.16	129	9600	0.00	119	9600	0.00	71	
10417	10417	0.00	191	10417	0.00	119	10378	-0.37	110	10473	0.53	65	
19.2k	19.23k	0.16	103	19.23k	0.16	64	19.20k	0.00	59	19.20k	0.00	35	
57.6k	57.14k	-0.79	34	56.818	-1.36	21	57.60k	0.00	19	57.60k	0.00	11	
115.2k	117.6k	2.12	16	113.636	-1.36	10	115.2k	0.00	9	115.2k	0.00	5	

					SYNC	C = 0, BRGH	l = 0, BRG	616 = 1				
BAUD	Fos	c = 8.000) MHz	Fos	c = 4.000) MHz	Fosc	: = 3.686	4 MHz	Fos	c = 1.000) MHz
RATE	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)
300	299.9	-0.02	1666	300.1	0.04	832	300.0	0.00	767	300.5	0.16	207
1200	1199	-0.08	416	1202	0.16	207	1200	0.00	191	1202	0.16	51
2400	2404	0.16	207	2404	0.16	103	2400	0.00	95	2404	0.16	25
9600	9615	0.16	51	9615	0.16	25	9600	0.00	23	_	_	—
10417	10417	0.00	47	10417	0.00	23	10473	0.53	21	10417	0.00	5
19.2k	19.23k	0.16	25	19.23k	0.16	12	19.20k	0.00	11	_	_	_
57.6k	55556	-3.55	8	—	—	_	57.60k	0.00	3	—	_	_
115.2k	—	_	_	_	_	_	115.2k	0.00	1	_	_	_

				SYNC = 0	, BRGH	= 1, BRG16	= 1 or SY	/NC = 1,	BRG16 = 1			
BAUD	Fosc	= 32.00	0 MHz	Fosc	= 20.00	0 MHz	Fosc	: = 18.43	2 MHz	Fosc	= 11.059	92 MHz
RATE	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)
300	300.0	0.00	26666	300.0	0.00	16665	300.0	0.00	15359	300.0	0.00	9215
1200	1200	0.00	6666	1200	-0.01	4166	1200	0.00	3839	1200	0.00	2303
2400	2400	0.01	3332	2400	0.02	2082	2400	0.00	1919	2400	0.00	1151
9600	9604	0.04	832	9597	-0.03	520	9600	0.00	479	9600	0.00	287
10417	10417	0.00	767	10417	0.00	479	10425	0.08	441	10433	0.16	264
19.2k	19.18k	-0.08	416	19.23k	0.16	259	19.20k	0.00	239	19.20k	0.00	143
57.6k	57.55k	-0.08	138	57.47k	-0.22	86	57.60k	0.00	79	57.60k	0.00	47
115.2k	115.9k	0.64	68	116.3k	0.94	42	115.2k	0.00	39	115.2k	0.00	23

TABLE 33-5: BAUD RATES FOR ASYNCHRONOUS MODES (CONTINUED)

		SYNC = 0, BRGH = 1, BRG16 = 1 or SYNC = 1, BRG16 = 1											
BAUD	Fos	c = 8.000) MHz	Fos	c = 4.000) MHz	Fosc	: = 3.686	4 MHz	Fos	Fosc = 1.000 MHz		
RATE	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)	Actual Rate	% Error	SPxBRG value (decimal)	
300	300.0	0.00	6666	300.0	0.01	3332	300.0	0.00	3071	300.1	0.04	832	
1200	1200	-0.02	1666	1200	0.04	832	1200	0.00	767	1202	0.16	207	
2400	2401	0.04	832	2398	0.08	416	2400	0.00	383	2404	0.16	103	
9600	9615	0.16	207	9615	0.16	103	9600	0.00	95	9615	0.16	25	
10417	10417	0	191	10417	0.00	95	10473	0.53	87	10417	0.00	23	
19.2k	19.23k	0.16	103	19.23k	0.16	51	19.20k	0.00	47	19.23k	0.16	12	
57.6k	57.14k	-0.79	34	58.82k	2.12	16	57.60k	0.00	15	—	_	—	
115.2k	117.6k	2.12	16	111.1k	-3.55	8	115.2k	0.00	7	_	_	—	

33.4.1 AUTO-BAUD DETECT

The EUSART module supports automatic detection and calibration of the baud rate.

In the Auto-Baud Detect (ABD) mode, the clock to the BRG is reversed. Rather than the BRG clocking the incoming RX signal, the RX signal is timing the BRG. The Baud Rate Generator is used to time the period of a received 55h (ASCII "U") which is the Sync character for the LIN bus. The unique feature of this character is that it has five rising edges including the Stop bit edge.

Setting the ABDEN bit of the BAUDxCON register starts the auto-baud calibration sequence. While the ABD sequence takes place, the EUSART state machine is held in Idle. On the first rising edge of the receive line, after the Start bit, the SPxBRG begins counting up using the BRG counter clock as shown in Figure 33-6. The fifth rising edge will occur on the RX pin at the end of the eighth bit period. At that time, an accumulated value totaling the proper BRG period is left in the SPxBRGH:SPxBRGL register pair, the ABDEN bit is automatically cleared and the RCIF interrupt flag is set. The value in the RCxREG needs to be read to clear the RCIF interrupt. RCxREG content should be discarded. When calibrating for modes that do not use the SPxBRGH register the user can verify that the SPxBRGL register did not overflow by checking for 00h in the SPxBRGH register.

The BRG auto-baud clock is determined by the BRG16 and BRGH bits as shown in Table 33-6. During ABD, both the SPxBRGH and SPxBRGL registers are used as a 16-bit counter, independent of the BRG16 bit setting. While calibrating the baud rate period, the SPxBRGH and SPxBRGL registers are clocked at 1/8th the BRG base clock rate. The resulting byte measurement is the average bit time when clocked at full speed.

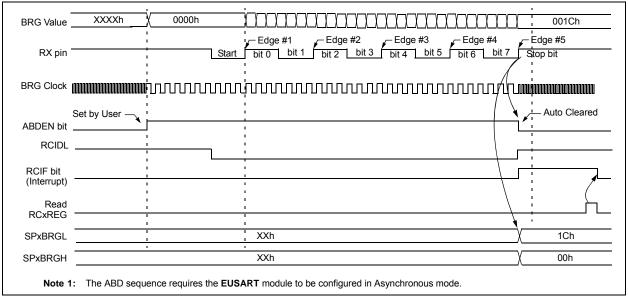
- Note 1: If the WUE bit is set with the ABDEN bit, auto-baud detection will occur on the byte <u>following</u> the Break character (see <u>Section 33.4.3</u> "Auto-Wake-up on Break").
 - 2: It is up to the user to determine that the incoming character baud rate is within the range of the selected BRG clock source. Some combinations of oscillator frequency and EUSART baud rates are not possible.
 - 3: During the auto-baud process, the auto-baud counter starts counting at one. Upon completion of the auto-baud sequence, to achieve maximum accuracy, subtract 1 from the SPxBRGH:SPxBRGL register pair.

TABLE 33-6: BRG COUNTER CLOCK RATES

BRG16	BRGH	BRG Base Clock	BRG ABD Clock
0	0	Fosc/64	Fosc/512
0	1	Fosc/16	Fosc/128
1	0	Fosc/16	Fosc/128
1	1	Fosc/4	Fosc/32

Note: During the ABD sequence, SPxBRGL and SPxBRGH registers are both used as a 16-bit counter, independent of the BRG16 setting.

FIGURE 33-6: AUTOMATIC BAUD RATE CALIBRATION⁽¹⁾



33.4.2 AUTO-BAUD OVERFLOW

During the course of automatic baud detection, the ABDOVF bit of the BAUDxCON register will be set if the baud rate counter overflows before the fifth rising edge is detected on the RX pin. The ABDOVF bit indicates that the counter has exceeded the maximum count that can fit in the 16 bits of the SPxBRGH:SPxBRGL register pair. The overflow condition will set the RCIF flag. The counter continues to count until the fifth rising edge is detected on the RX pin. The RCIDL bit will remain false ('0') until the fifth rising edge at which time the RCIDL bit will be set. If the RCREG is read after the overflow occurs but before the fifth rising edge then the fifth rising edge will set the RCIF again.

Terminating the auto-baud process early to clear an overflow condition will prevent proper detection of the sync character fifth rising edge. If any falling edges of the sync character have not yet occurred when the ABDEN bit is cleared then those will be falsely detected as Start bits. The following steps are recommended to clear the overflow condition:

- 1. Read RCREG to clear RCIF.
- 2. If RCIDL is zero then wait for RCIF and repeat step 1.
- 3. Clear the ABDOVF bit.

33.4.3 AUTO-WAKE-UP ON BREAK

During Sleep mode, all clocks to the EUSART are suspended. Because of this, the Baud Rate Generator is inactive and a proper character reception cannot be performed. The Auto-Wake-up feature allows the controller to wake-up due to activity on the RX/DT line. This feature is available only in Asynchronous mode.

The Auto-Wake-up feature is enabled by setting the WUE bit of the BAUDxCON register. Once set, the normal receive sequence on RX/DT is disabled, and the EUSART remains in an Idle state, monitoring for a wake-up event independent of the CPU mode. A wake-up event consists of a high-to-low transition on the RX/DT line. (This coincides with the start of a Sync Break or a wake-up signal character for the LIN protocol.)

The EUSART module generates an RCIF interrupt coincident with the wake-up event. The interrupt is generated synchronously to the Q clocks in normal CPU operating modes (Figure 33-7), and asynchronously if the device is in Sleep mode (Figure 33-8). The interrupt condition is cleared by reading the RCxREG register.

The WUE bit is automatically cleared by the low-to-high transition on the RX line at the end of the Break. This signals to the user that the Break event is over. At this point, the EUSART module is in Idle mode waiting to receive the next character.

33.4.3.1 Special Considerations

Break Character

To avoid character errors or character fragments during a wake-up event, the wake-up character must be all zeros.

When the wake-up is enabled the function works independent of the low time on the data stream. If the WUE bit is set and a valid non-zero character is received, the low time from the Start bit to the first rising edge will be interpreted as the wake-up event. The remaining bits in the character will be received as a fragmented character and subsequent characters can result in framing or overrun errors.

Therefore, the initial character in the transmission must be all '0's. This must be ten or more bit times, 13-bit times recommended for LIN bus, or any number of bit times for standard RS-232 devices.

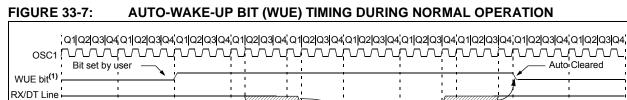
Oscillator Start-up Time

Oscillator start-up time must be considered, especially in applications using oscillators with longer start-up intervals (i.e., LP, XT or HS/PLL mode). The Sync Break (or wake-up signal) character must be of sufficient length, and be followed by a sufficient interval, to allow enough time for the selected oscillator to start and provide proper initialization of the EUSART.

WUE Bit

The wake-up event causes a receive interrupt by setting the RCIF bit. The WUE bit is cleared in hardware by a rising edge on RX/DT. The interrupt condition is then cleared in software by reading the RCxREG register and discarding its contents.

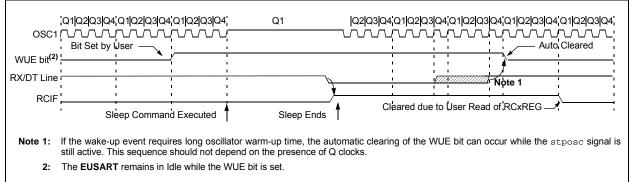
To ensure that no actual data is lost, check the RCIDL bit to verify that a receive operation is not in process before setting the WUE bit. If a receive operation is not occurring, the WUE bit may then be set just prior to entering the Sleep mode.



ı. ì ٠ Ì Cleared due to User Read of RCxREG -Note 1: The EUSART remains in Idle while the WUE bit is set.

FIGURE 33-8: AUTO-WAKE-UP BIT (WUE) TIMINGS DURING SLEEP

RCIF



33.4.4 BREAK CHARACTER SEQUENCE

The EUSART module has the capability of sending the special Break character sequences that are required by the LIN bus standard. A Break character consists of a Start bit, followed by 12 '0' bits and a Stop bit.

To send a Break character, set the SENDB and TXEN bits of the TXxSTA register. The Break character transmission is then initiated by a write to the TXxREG. The value of data written to TXxREG will be ignored and all '0's will be transmitted.

The SENDB bit is automatically reset by hardware after the corresponding Stop bit is sent. This allows the user to preload the transmit FIFO with the next transmit byte following the Break character (typically, the Sync character in the LIN specification).

The TRMT bit of the TXxSTA register indicates when the transmit operation is active or idle, just as it does during normal transmission. See Figure 33-9 for the timing of the Break character sequence.

33.4.4.1 Break and Sync Transmit Sequence

The following sequence will start a message frame header made up of a Break, followed by an auto-baud Sync byte. This sequence is typical of a LIN bus master.

- 1. Configure the EUSART for the desired mode.
- 2. Set the TXEN and SENDB bits to enable the Break sequence.
- 3. Load the TXxREG with a dummy character to initiate transmission (the value is ignored).
- 4. Write '55h' to TXxREG to load the Sync character into the transmit FIFO buffer.
- 5. After the Break has been sent, the SENDB bit is reset by hardware and the Sync character is then transmitted.

When the TXxREG becomes empty, as indicated by the TXIF, the next data byte can be written to TXxREG.

Write to TXxREG Dummy Write **BRG** Output (Shift Clock) TX (pin) Start bit bit 0 bit 1 Stop bit Break TXIF bit (Transmit Interrupt Flag) TRMT bit (Transmit Shift Empty Flag) SENDB Sampled Here Auto Cleared SENDB (send Break control bit)

33.4.5 RECEIVING A BREAK CHARACTER

The Enhanced EUSART module can receive a Break character in two ways.

The first method to detect a Break character uses the FERR bit of the RCxSTA register and the received data as indicated by RCxREG. The Baud Rate Generator is assumed to have been initialized to the expected baud rate.

A Break character has been received when:

- RCIF bit is set
- · FERR bit is set
- RCxREG = 00h

The second method uses the Auto-Wake-up feature described in **Section 33.4.3** "Auto-Wake-up on **Break**". By enabling this feature, the EUSART will sample the next two transitions on RX/DT, cause an RCIF interrupt, and receive the next data byte followed by another interrupt.

Note that following a Break character, the user will typically want to enable the Auto-Baud Detect feature. For both methods, the user can set the ABDEN bit of the BAUDxCON register before placing the EUSART in Sleep mode.

FIGURE 33-9: SEND BREAK CHARACTER SEQUENCE

33.5 EUSART Synchronous Mode

Synchronous serial communications are typically used in systems with a single master and one or more slaves. The master device contains the necessary circuitry for baud rate generation and supplies the clock for all devices in the system. Slave devices can take advantage of the master clock by eliminating the internal clock generation circuitry.

There are two signal lines in Synchronous mode: a bidirectional data line and a clock line. Slaves use the external clock supplied by the master to shift the serial data into and out of their respective receive and transmit shift registers. Since the data line is bidirectional, synchronous operation is half-duplex only. Half-duplex refers to the fact that master and slave devices can receive and transmit data but not both simultaneously. The EUSART can operate as either a master or slave device.

Start and Stop bits are not used in synchronous transmissions.

33.5.1 SYNCHRONOUS MASTER MODE

The following bits are used to configure the EUSART for synchronous master operation:

- SYNC = 1
- CSRC = 1
- SREN = 0 (for transmit); SREN = 1 (for receive)
- CREN = 0 (for transmit); CREN = 1 (for receive)
- SPEN = 1

Setting the SYNC bit of the TXxSTA register configures the device for synchronous operation. Setting the CSRC bit of the TXxSTA register configures the device as a master. Clearing the SREN and CREN bits of the RCxSTA register ensures that the device is in the Transmit mode, otherwise the device will be configured to receive. Setting the SPEN bit of the RCxSTA register enables the EUSART.

33.5.1.1 Master Clock

Synchronous data transfers use a separate clock line, which is synchronous with the data. A device configured as a master transmits the clock on the TX/CK line. The TX/CK pin output driver is automatically enabled when the EUSART is configured for synchronous transmit or receive operation. Serial data bits change on the leading edge to ensure they are valid at the trailing edge of each clock. One clock cycle is generated for each data bit. Only as many clock cycles are generated as there are data bits.

33.5.1.2 Clock Polarity

A clock polarity option is provided for Microwire compatibility. Clock polarity is selected with the SCKP bit of the BAUDxCON register. Setting the SCKP bit sets the clock Idle state as high. When the SCKP bit is set, the data changes on the falling edge of each clock. Clearing the SCKP bit sets the Idle state as low. When the SCKP bit is cleared, the data changes on the rising edge of each clock.

33.5.1.3 Synchronous Master Transmission

Data is transferred out of the device on the RX/DT pin. The RX/DT and TX/CK pin output drivers are automatically enabled when the EUSART is configured for synchronous master transmit operation.

A transmission is initiated by writing a character to the TXxREG register. If the TSR still contains all or part of a previous character the new character data is held in the TXxREG until the last bit of the previous character has been transmitted. If this is the first character, or the previous character has been completely flushed from the TSR, the data in the TXxREG is immediately transferred to the TSR. The transmission of the character commences immediately following the transfer of the data to the TSR from the TXxREG.

Each data bit changes on the leading edge of the master clock and remains valid until the subsequent leading clock edge.

Note: The TSR register is not mapped in data memory, so it is not available to the user.

- 33.5.1.4 Synchronous Master Transmission Set-up:
- Initialize the SPxBRGH:SPxBRGL register pair and the BRGH and BRG16 bits to achieve the desired baud rate (see Section 33.4 "EUSART Baud Rate Generator (BRG)").
- 2. Enable the synchronous master serial port by setting bits SYNC, SPEN and CSRC.
- 3. Disable Receive mode by clearing bits SREN and CREN.
- 4. Enable Transmit mode by setting the TXEN bit.
- 5. If 9-bit transmission is desired, set the TX9 bit.
- If interrupts are desired, set the TXIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
- 7. If 9-bit transmission is selected, the ninth bit should be loaded in the TX9D bit.
- 8. Start transmission by loading data to the TXxREG register.

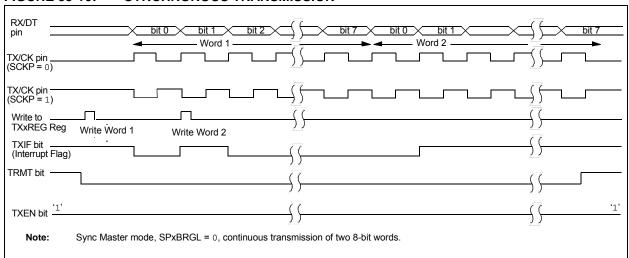


FIGURE 33-10: SYNCHRONOUS TRANSMISSION

FIGURE 33-11: SYNCHRONOUS TRANSMISSION (THROUGH TXEN)

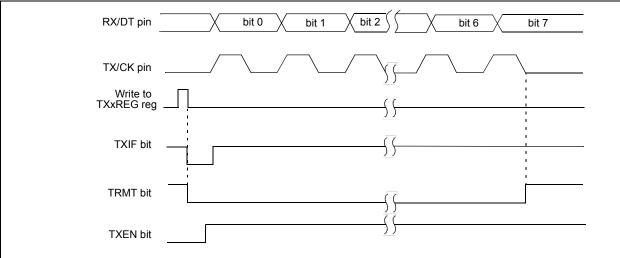


TABLE 33-7:	SUMMARY OF REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER
	TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page		
ANSELA	-		ANSA5	ANSA4	ANSA3	ANSA2	ANSA2 ANSA1 ANSA		177		
ANSELB	_	_	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	182		
ANSELC	ANSC7	ANSC6	ANSC5	ANSC4	ANSC3	ANSC2		—	187		
BAUD1CON	ABDOVF	RCIDL	-	SCKP	BRG16	—	WUE	ABDEN	505		
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	132		
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	CCP1IE TMR2IE		IE CCP1IE TMR2		TMR1IE	133
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	CCP1IF	TMR2IF	TMR1IF	139		
RC1STA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	504		
RxyPPS	_	_			RxyPP	S<5:0>			205		
SP1BRGL				SP1BR0	G<7:0>				506		
SP1BRGH				SP1BR0	6<15:8>				506		
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	176		
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	181		
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	RISC2 TRISC1 TRISC0		186		
TX1REG			EUS	ART Transm	it Data Regis	ster			495*		
TX1STA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	503		

Legend: *

- = unimplemented location, read as '0'. Shaded cells are not used for synchronous master transmission.

Page provides register information.

33.5.1.5 Synchronous Master Reception

Data is received at the RX/DT pin. The RX/DT pin output driver is automatically disabled when the EUSART is configured for synchronous master receive operation.

In Synchronous mode, reception is enabled by setting either the Single Receive Enable bit (SREN of the RCxSTA register) or the Continuous Receive Enable bit (CREN of the RCxSTA register).

When SREN is set and CREN is clear, only as many clock cycles are generated as there are data bits in a single character. The SREN bit is automatically cleared at the completion of one character. When CREN is set, clocks are continuously generated until CREN is cleared. If CREN is cleared in the middle of a character the CK clock stops immediately and the partial character is discarded. If SREN and CREN are both set, then SREN is cleared at the completion of the first character and CREN takes precedence.

To initiate reception, set either SREN or CREN. Data is sampled at the RX/DT pin on the trailing edge of the TX/CK clock pin and is shifted into the Receive Shift Register (RSR). When a complete character is received into the RSR, the RCIF bit is set and the character is automatically transferred to the two character receive FIFO. The Least Significant eight bits of the top character in the receive FIFO are available in RCxREG. The RCIF bit remains set as long as there are unread characters in the receive FIFO.

Note:	If the RX/DT function is on an analog pin,								
	the corresponding ANSEL bit must be								
	cleared for the receiver to function.								

33.5.1.6 Slave Clock

Synchronous data transfers use a separate clock line, which is synchronous with the data. A device configured as a slave receives the clock on the TX/CK line. The TX/CK pin output driver is automatically disabled when the device is configured for synchronous slave transmit or receive operation. Serial data bits change on the leading edge to ensure they are valid at the trailing edge of each clock. One data bit is transferred for each clock cycle. Only as many clock cycles should be received as there are data bits.

Note: If the device is configured as a slave and the TX/CK function is on an analog pin, the corresponding ANSEL bit must be cleared.

33.5.1.7 Receive Overrun Error

The receive FIFO buffer can hold two characters. An overrun error will be generated if a third character, in its entirety, is received before RCxREG is read to access the FIFO. When this happens the OERR bit of the RCxSTA register is set. Previous data in the FIFO will not be overwritten. The two characters in the FIFO buffer can be read, however, no additional characters will be received until the error is cleared. The OERR bit can only be cleared by clearing the overrun condition. If the overrun error occurred when the SREN bit is set and CREN is clear then the error is cleared by reading RCxREG. If the overrun occurred when the CREN bit is set then the error condition is cleared by either clearing the CREN bit of the RCxSTA register or by clearing the SPEN bit which resets the EUSART.

33.5.1.8 Receiving 9-bit Characters

The EUSART supports 9-bit character reception. When the RX9 bit of the RCxSTA register is set, the EUSART will shift nine bits into the RSR for each character received. The RX9D bit of the RCxSTA register is the ninth, and Most Significant, data bit of the top unread character in the receive FIFO. When reading 9-bit data from the receive FIFO buffer, the RX9D data bit must be read before reading the eight Least Significant bits from the RCxREG.

33.5.1.9 Synchronous Master Reception Set-up:

- 1. Initialize the SPxBRGH:SPxBRGL register pair for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
- 2. Clear the ANSEL bit for the RX pin (if applicable).
- 3. Enable the synchronous master serial port by setting bits SYNC, SPEN and CSRC.
- 4. Ensure bits CREN and SREN are clear.
- 5. If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
- 6. If 9-bit reception is desired, set bit RX9.
- 7. Start reception by setting the SREN bit or for continuous reception, set the CREN bit.
- 8. Interrupt flag bit RCIF will be set when reception of a character is complete. An interrupt will be generated if the enable bit RCIE was set.
- 9. Read the RCxSTA register to get the ninth bit (if enabled) and determine if any error occurred during reception.
- 10. Read the 8-bit received data by reading the RCxREG register.
- 11. If an overrun error occurs, clear the error by either clearing the CREN bit of the RCxSTA register or by clearing the SPEN bit which resets the EUSART.

FIGURE 33-12:	SYNCHRONOUS RECEPTION (MASTER MODE, SREN)
RX/DT pin TX/CK pin (SCKP = 0)	X bit 0 bit 2 bit 3 bit 4 bit 5 bit 6 bit 7
TX/CK pin	
SREN bit	·0'
RCIF bit (Interrupt) ——— Read RCxREG————	
	gram demonstrates Sync Master mode with bit SREN = 1 and bit BRGH = 0 .

TABLE 33-8: SUMMARY OF REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER RECEPTION RECEPTION

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page		
_	—	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0	177		
_	—	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	182		
ANSC7	ANSC6	ANSC5	ANSC4	ANSC3	ANSC2	_	_	187		
ABDOVF	RCIDL	—	SCKP	BRG16	—	WUE	ABDEN	505		
_	—		CKPPS<5:0>							
GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	132		
TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	CCP1IE	TMR2IE	TMR1IE	133		
TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	CCP1IF	TMR2IF	TMR1IF	139		
		EUS	SART Receiv	e Data Regis	ter			498*		
SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	504		
_	—			RXPP	S<5:0>			205, 207		
_	—			RxyPP	S<5:0>			205		
			SP1BR0	G<7:0>				506*		
			SP1BRG	6<15:8>				506*		
TRISA5	TRISA4	TRISA5	TRISA4	TRISA5	TRISA2	TRISA1	TRISA0	176		
TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	181		
TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	186		
CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	503		
	ANSC7 ABDOVF GIE TMR1GIE TMR1GIF SPEN 	ANSC7 ANSC6 ABDOVF RCIDL GIE PEIE TMR1GIE ADIE TMR1GIF ADIF SPEN RX9 TRISA5 TRISA4 TRISB7 TRISB6 TRISC6 TRISC6	Image: Sector of the sector	Image <td>Image<t< td=""><td>ImageImageImageImageImageANSA5ANSA4ANSA3ANSA2ANSB5ANSB4ANSB3ANSB2ANSC7ANSC6ANSC5ANSC4ANSC3ANSC2ABDOVFRCIDLSCKPBRG16SCKPBRG16SCKPBRG16SCKPBRG16GIEPEIETMR0IEINTEIOCIETMR0IFTMR1GIEADIERCIETXIESSP1IECCP1IETMR1GIFADIFRCIFTXIFSSP1IFCCP1IFSPENRX9SRENCRENADDENFERRRX9PSRENCRENADDENSP1BRSP1BRSP1BRSP1BRRX9PSRENITRISA5TRISA5TRISA4TRISA5TRISA4TRISA5TRISA5TRISA5TRISA4TRISA5TRISA4TRISA5TRISA2TRISB7TRISB6TRISS5TRISC4TRISC3TRISC2</td><td>Image: Constant of the stateImage: /td><td>Image: Constant of the stateImage: /td></t<></td>	Image <t< td=""><td>ImageImageImageImageImageANSA5ANSA4ANSA3ANSA2ANSB5ANSB4ANSB3ANSB2ANSC7ANSC6ANSC5ANSC4ANSC3ANSC2ABDOVFRCIDLSCKPBRG16SCKPBRG16SCKPBRG16SCKPBRG16GIEPEIETMR0IEINTEIOCIETMR0IFTMR1GIEADIERCIETXIESSP1IECCP1IETMR1GIFADIFRCIFTXIFSSP1IFCCP1IFSPENRX9SRENCRENADDENFERRRX9PSRENCRENADDENSP1BRSP1BRSP1BRSP1BRRX9PSRENITRISA5TRISA5TRISA4TRISA5TRISA4TRISA5TRISA5TRISA5TRISA4TRISA5TRISA4TRISA5TRISA2TRISB7TRISB6TRISS5TRISC4TRISC3TRISC2</td><td>Image: Constant of the stateImage: /td><td>Image: Constant of the stateImage: /td></t<>	ImageImageImageImageImageANSA5ANSA4ANSA3ANSA2ANSB5ANSB4ANSB3ANSB2ANSC7ANSC6ANSC5ANSC4ANSC3ANSC2ABDOVFRCIDLSCKPBRG16SCKPBRG16SCKPBRG16SCKPBRG16GIEPEIETMR0IEINTEIOCIETMR0IFTMR1GIEADIERCIETXIESSP1IECCP1IETMR1GIFADIFRCIFTXIFSSP1IFCCP1IFSPENRX9SRENCRENADDENFERRRX9PSRENCRENADDENSP1BRSP1BRSP1BRSP1BRRX9PSRENITRISA5TRISA5TRISA4TRISA5TRISA4TRISA5TRISA5TRISA5TRISA4TRISA5TRISA4TRISA5TRISA2TRISB7TRISB6TRISS5TRISC4TRISC3TRISC2	Image: Constant of the stateImage:	Image: Constant of the stateImage:		

Legend: — = unimplemented location, read as '0'. Shaded cells are not used for synchronous master reception.

* Page provides register information.

33.5.2 SYNCHRONOUS SLAVE MODE

The following bits are used to configure the EUSART for synchronous slave operation:

- SYNC = 1
- CSRC = 0
- SREN = 0 (for transmit); SREN = 1 (for receive)
- CREN = 0 (for transmit); CREN = 1 (for receive)
- SPEN = 1

Setting the SYNC bit of the TXxSTA register configures the device for synchronous operation. Clearing the CSRC bit of the TXxSTA register configures the device as a slave. Clearing the SREN and CREN bits of the RCxSTA register ensures that the device is in the Transmit mode, otherwise the device will be configured to receive. Setting the SPEN bit of the RCxSTA register enables the EUSART.

33.5.2.1 EUSART Synchronous Slave Transmit

The operation of the Synchronous Master and Slave modes are identical (see Section 33.5.1.3 "Synchronous Master Transmission"), except in the case of the Sleep mode.

If two words are written to the TXxREG and then the SLEEP instruction is executed, the following will occur:

- 1. The first character will immediately transfer to the TSR register and transmit.
- 2. The second word will remain in the TXxREG register.
- 3. The TXIF bit will not be set.
- After the first character has been shifted out of TSR, the TXxREG register will transfer the second character to the TSR and the TXIF bit will now be set.
- 5. If the PEIE and TXIE bits are set, the interrupt will wake the device from Sleep and execute the next instruction. If the GIE bit is also set, the program will call the Interrupt Service Routine.

- 33.5.2.2 Synchronous Slave Transmission Set-up:
- 1. Set the SYNC and SPEN bits and clear the CSRC bit.
- 2. Clear the ANSEL bit for the CK pin (if applicable).
- 3. Clear the CREN and SREN bits.
- If interrupts are desired, set the TXIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
- 5. If 9-bit transmission is desired, set the TX9 bit.
- 6. Enable transmission by setting the TXEN bit.
- 7. If 9-bit transmission is selected, insert the Most Significant bit into the TX9D bit.
- 8. Start transmission by writing the Least Significant eight bits to the TXxREG register.

TABLE 33-9:	SUMMARY OF REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE
	TRANSMISSION

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page	
_		ANSA5	ANSA4	ANSA3	ANSA2 ANSA1		ANSA0	177	
_	_	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	182	
ANSC7	ANSC6	ANSC5	ANSC4	ANSC3	ANSC2	_	—	187	
ABDOVF	RCIDL	-	SCKP	BRG16	_	WUE	ABDEN	505	
—	—		205, 207						
GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	132	
TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	CCP1IE	TMR2IE	TMR1IE	133	
TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	CCP1IF	TMR2IF	TMR1IF	139	
SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	504	
_	_			RXPP	S<5:0>			205, 207	
_	_			RxyPP	S<5:0>			205	
TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	176	
TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	181	
TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	186	
		EUS	ART Transm	it Data Regis	ster			495*	
CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	503	
		ANSC7 ANSC6 ABDOVF RCIDL GIE PEIE TMR1GIE ADIE TMR1GIF ADIF SPEN RX9 TRISA7 TRISA6 TRISB7 TRISE6 TRISC6	Image: market intermed and interme	Image: constraint of the sector of the sec	Image: constraint of the sector of the sec	Image: state	Image: constraint of the sector of the sec	Image: constraint of the sector of the se	

Legend:

- = unimplemented location, read as '0'. Shaded cells are not used for synchronous slave transmission.

* Page provides register information.

33.5.2.3 EUSART Synchronous Slave Reception

The operation of the Synchronous Master and Slave modes is identical (Section 33.5.1.5 "Synchronous Master Reception"), with the following exceptions:

- Sleep
- CREN bit is always set, therefore the receiver is never idle
- SREN bit, which is a "don't care" in Slave mode

A character may be received while in Sleep mode by setting the CREN bit prior to entering Sleep. Once the word is received, the RSR register will transfer the data to the RCxREG register. If the RCIE enable bit is set, the interrupt generated will wake the device from Sleep and execute the next instruction. If the GIE bit is also set, the program will branch to the interrupt vector.

- 33.5.2.4 Synchronous Slave Reception Set-up:
- 1. Set the SYNC and SPEN bits and clear the CSRC bit.
- 2. Clear the ANSEL bit for both the CK and DT pins (if applicable).
- 3. If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
- 4. If 9-bit reception is desired, set the RX9 bit.
- 5. Set the CREN bit to enable reception.
- The RCIF bit will be set when reception is complete. An interrupt will be generated if the RCIE bit was set.
- 7. If 9-bit mode is enabled, retrieve the Most Significant bit from the RX9D bit of the RCxSTA register.
- 8. Retrieve the eight Least Significant bits from the receive FIFO by reading the RCxREG register.
- 9. If an overrun error occurs, clear the error by either clearing the CREN bit of the RCxSTA register or by clearing the SPEN bit which resets the EUSART.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ANSELA	—	—	ANSA5	ANSA4	ANSA3	ANSA2	ANSA1	ANSA0	177
ANSELB	—	_	ANSB5	ANSB4	ANSB3	ANSB2	ANSB1	ANSB0	182
ANSELC	ANSC7	ANSC6	ANSC5	ANSC4	ANSC3	ANSC2		—	187
BAUD1CON	ABDOVF	RCIDL	_	SCKP	BRG16	-	WUE	ABDEN	505
CKPPS	—	—		205, 207					
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF INTF		IOCIF	132
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	CCP1IE	TMR2IE	TMR1IE	133
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	CCP1IF	TMR2IF	TMR1IF	139
RC1REG			EUS	SART Receiv	e Data Regis	ter			498*
RC1STA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	504
RXPPS	—	—			RXPP	S<5:0>			205, 207
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	176
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	181
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	186
TX1STA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	503

TABLE 33-10: SUMMARY OF REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE RECEPTION

Legend: — = unimplemented location, read as '0'. Shaded cells are not used for synchronous slave reception.
* Page provides register information.

33.6 EUSART Operation During Sleep

The EUSART will remain active during Sleep only in the Synchronous Slave mode. All other modes require the system clock and therefore cannot generate the necessary signals to run the Transmit or Receive Shift registers during Sleep.

Synchronous Slave mode uses an externally generated clock to run the Transmit and Receive Shift registers.

33.6.1 SYNCHRONOUS RECEIVE DURING SLEEP

To receive during Sleep, all the following conditions must be met before entering Sleep mode:

- RCxSTA and TXxSTA Control registers must be configured for Synchronous Slave Reception (see Section 33.5.2.4 "Synchronous Slave Reception Set-up:").
- If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
- The RCIF interrupt flag must be cleared by reading RCxREG to unload any pending characters in the receive buffer.

Upon entering Sleep mode, the device will be ready to accept data and clocks on the RX/DT and TX/CK pins, respectively. When the data word has been completely clocked in by the external device, the RCIF interrupt flag bit of the PIR1 register will be set. Thereby, waking the processor from Sleep.

Upon waking from Sleep, the instruction following the SLEEP instruction will be executed. If the Global Interrupt Enable (GIE) bit of the INTCON register is also set, then the Interrupt Service Routine at address 004h will be called.

33.6.2 SYNCHRONOUS TRANSMIT DURING SLEEP

To transmit during Sleep, all the following conditions must be met before entering Sleep mode:

- The RCxSTA and TXxSTA Control registers must be configured for synchronous slave transmission (see Section 33.5.2.2 "Synchronous Slave Transmission Set-up:").
- The TXIF interrupt flag must be cleared by writing the output data to the TXxREG, thereby filling the TSR and transmit buffer.
- If interrupts are desired, set the TXIE bit of the PIE1 register and the PEIE bit of the INTCON register.
- Interrupt enable bits TXIE of the PIE1 register and PEIE of the INTCON register must set.

Upon entering Sleep mode, the device will be ready to accept clocks on TX/CK pin and transmit data on the RX/DT pin. When the data word in the TSR has been completely clocked out by the external device, the pending byte in the TXxREG will transfer to the TSR and the TXIF flag will be set. Thereby, waking the processor from Sleep. At this point, the TXxREG is available to accept another character for transmission, which will clear the TXIF flag.

Upon waking from Sleep, the instruction following the SLEEP instruction will be executed. If the Global Interrupt Enable (GIE) bit is also set then the Interrupt Service Routine at address 0004h will be called.

34.0 IN-CIRCUIT SERIAL PROGRAMMING[™] (ICSP[™])

ICSP[™] programming allows customers to manufacture circuit boards with unprogrammed devices. Programming can be done after the assembly process, allowing the device to be programmed with the most recent firmware or a custom firmware. Five pins are needed for ICSP[™] programming:

- ICSPCLK
- ICSPDAT
- MCLR/VPP
- VDD
- Vss

In Program/Verify mode the program memory, User IDs and the Configuration Words are programmed through serial communications. The ICSPDAT pin is a bidirectional I/O used for transferring the serial data and the ICSPCLK pin is the clock input. For more information on ICSP™ refer to the "*PIC16(L)F177X Memory Programming Specification*" (DS40001792).

34.1 High-Voltage Programming Entry Mode

The device is placed into High-Voltage Programming Entry mode by holding the ICSPCLK and ICSPDAT pins low then raising the voltage on MCLR/VPP to VIHH.

34.2 Low-Voltage Programming Entry Mode

The Low-Voltage Programming Entry mode allows the PIC[®] Flash MCUs to be programmed using VDD only, without high voltage. When the LVP bit of Configuration Words is set to '1', the low-voltage ICSP programming entry is enabled. To disable the Low-Voltage ICSP mode, the LVP bit must be programmed to '0'.

Entry into the Low-Voltage Programming Entry mode requires the following steps:

- 1. MCLR is brought to VIL.
- 2. A 32-bit key sequence is presented on ICSPDAT, while clocking ICSPCLK.

Once the key sequence is complete, $\overline{\text{MCLR}}$ must be held at VIL for as long as Program/Verify mode is to be maintained.

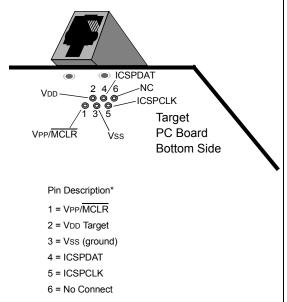
If low-voltage programming is enabled (LVP = 1), the $\overline{\text{MCLR}}$ Reset function is automatically enabled and cannot be disabled. See **Section 6.5 "MCLR"** for more information.

The LVP bit can only be reprogrammed to '0' by using the High-Voltage Programming mode.

34.3 Common Programming Interfaces

Connection to a target device is typically done through an ICSP[™] header. A commonly found connector on development tools is the RJ-11 in the 6P6C (6-pin, 6-connector) configuration. See Figure 34-1.



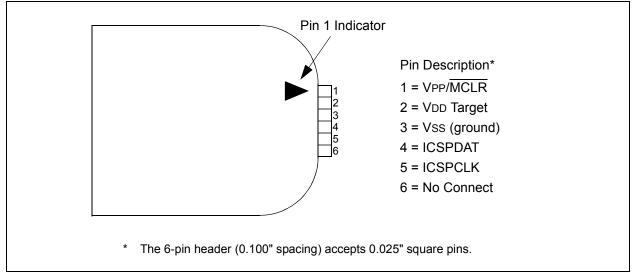


Another connector often found in use with the PICkit[™] programmers is a standard 6-pin header with 0.1 inch spacing. Refer to Figure 34-2.

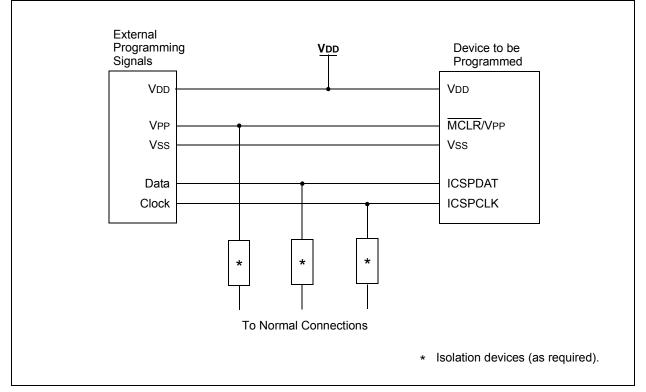
For additional interface recommendations, refer to your specific device programmer manual prior to PCB design.

It is recommended that isolation devices be used to separate the programming pins from other circuitry. The type of isolation is highly dependent on the specific application and may include devices such as resistors, diodes, or even jumpers. See Figure 34-3 for more information.









35.0 INSTRUCTION SET SUMMARY

Each instruction is a 14-bit word containing the operation code (opcode) and all required operands. The opcodes are broken into three broad categories.

- · Byte Oriented
- · Bit Oriented
- · Literal and Control

The literal and control category contains the most varied instruction word format.

Table 35-3 lists the instructions recognized by the MPASMTM assembler.

All instructions are executed within a single instruction cycle, with the following exceptions, which may take two or three cycles:

- Subroutine takes two cycles (CALL, CALLW)
- Returns from interrupts or subroutines take two cycles (RETURN, RETLW, RETFIE)
- Program branching takes two cycles (GOTO, BRA, BRW, BTFSS, BTFSC, DECFSZ, INCSFZ)
- One additional instruction cycle will be used when any instruction references an indirect file register and the file select register is pointing to program memory.

One instruction cycle consists of four oscillator cycles; for an oscillator frequency of 4 MHz, this gives a nominal instruction execution rate of 1 MHz.

All instruction examples use the format '0xhh' to represent a hexadecimal number, where 'h' signifies a hexadecimal digit.

35.1 Read-Modify-Write Operations

Any instruction that specifies a file register as part of the instruction performs a Read-Modify-Write (R-M-W) operation. The register is read, the data is modified, and the result is stored according to either the instruction, or the destination designator 'd'. A read operation is performed on a register even if the instruction writes to that register.

TABLE 35-1: OPCODE FIELD DESCRIPTIONS

Field	Description
f	Register file address (0x00 to 0x7F)
W	Working register (accumulator)
b	Bit address within an 8-bit file register
k	Literal field, constant data or label
x	Don't care location (= 0 or 1). The assembler will generate code with x = 0 . It is the recommended form of use for compatibility with all Microchip software tools.
d	Destination select; d = 0: store result in W, d = 1: store result in file register f. Default is d = 1.
n	FSR or INDF number. (0-1)
mm	Pre-post increment-decrement mode selection

TABLE 35-2: ABBREVIATION DESCRIPTIONS

Field	Description				
PC	Program Counter				
TO	Time-Out bit				
С	Carry bit				
DC	Digit Carry bit				
Z	Zero bit				
PD	Power-Down bit				

FIGURE 35-1: GENERAL FORMAT FOR INSTRUCTIONS

Byte-oriented file register op 13 8 7 6	erations 0
OPCODE d	f (FILE #)
d = 0 for destination W d = 1 for destination f f = 7-bit file register addre	
Bit-oriented file register oper 13 10 9	ations 7 6 0
OPCODE b (BIT a	
b = 3-bit bit address f = 7-bit file register addre	ss
Literal and control operation	S
General	
13 8 7	-
OPCODE	k (literal)
k = 8-bit immediate value	
CALL and GOTO instructions on	ly
13 11 10	0
OPCODE	(literal)
k = 11-bit immediate value	9
13 7	6 0
OPCODE	k (literal)
k = 7-bit immediate value	<u> </u>
MOVLB instruction only	
13	54 0
OPCODE	k (literal)
k = 5-bit immediate value	
BRA instruction only 13 9 8	0
OPCODE	k (literal)
k = 9-bit immediate value	. ,
FSR Offset instructions	
13 7 6 OPCODE n	5 0 k (literal)
n = appropriate FSR	K (illeral)
k = 6-bit immediate value	
FSR Increment instructions	3 2 1 0
OPCODE	n m (mode)
n = appropriate FSR m = 2-bit mode value	
OPCODE only 13	0
OPCODE	Ξ

Mnem	nonic,	Description	Circles	14-Bit Opcode				Status	Natas
Operands		Description	Cycles	MSb			LSb	Affected	Notes
		BYTE-ORIENTED FILE	REGISTER OPE	RATIC	NS				
ADDWF	f, d	Add W and f	1	00	0111	dfff	ffff	C, DC, Z	2
ADDWFC	f, d	Add with Carry W and f	1	11	1101	dfff	ffff	C, DC, Z	2
ANDWF	f, d	AND W with f	1	00	0101	dfff	ffff	Z	2
ASRF	f, d	Arithmetic Right Shift	1	11	0111	dfff	ffff	C, Z	2
LSLF	f, d	Logical Left Shift	1	11	0101	dfff	ffff	C, Z	2
LSRF	f, d	Logical Right Shift	1	11	0110	dfff	ffff	C, Z	2
CLRF	f	Clear f	1	00	0001	lfff	ffff	Z	2
CLRW	-	Clear W	1	00	0001	0000	00xx	Z	
COMF	f, d	Complement f	1	00	1001	dfff	ffff	Z	2
DECF	f, d	Decrement f	1	00	0011	dfff	ffff	Z	2
INCF	f, d	Increment f	1	00	1010	dfff	ffff	Z	2
IORWF	f, d	Inclusive OR W with f	1	00	0100	dfff	ffff	Z	2
MOVF	f, d	Move f	1	00	1000	dfff	ffff	Z	2
MOVWF	f	Move W to f	1	00	0000	lfff	ffff		2
RLF	f, d	Rotate Left f through Carry	1	00	1101	dfff	ffff	С	2
RRF	f, d	Rotate Right f through Carry	1	00	1100	dfff	ffff	С	2
SUBWF	f, d	Subtract W from f	1	00	0010	dfff	ffff	C, DC, Z	2
SUBWFB	f, d	Subtract with Borrow W from f	1	11	1011	dfff	ffff	C, DC, Z	2
SWAPF	f, d	Swap nibbles in f	1	00	1110	dfff	ffff		2
XORWF	f, d	Exclusive OR W with f	1	00	0110	dfff	ffff	Z	2
		BYTE ORIENTED	SKIP OPERATIO	ONS					
DECFSZ	f, d	Decrement f, Skip if 0	1(2)	00	1011	dfff	ffff		1, 2
INCFSZ	f, d	Increment f, Skip if 0	1(2)	00	1111	dfff	ffff		1, 2
		BIT-ORIENTED FILE R		RATION	IS				
BCF	f, b	Bit Clear f	1	01	00bb	bfff	ffff		2
BSF	f, b	Bit Set f	1	01	01bb	bfff	ffff		2
		BIT-ORIENTED S	KIP OPERATIO	NS					
BTFSC	f, b	Bit Test f, Skip if Clear	1 (2)	01	10bb	bfff	ffff		1, 2
BTFSS	f, b	Bit Test f, Skip if Set	1 (2)	01	11bb	bfff	ffff		1, 2
			PERATIONS						
ADDLW	k	Add literal and W	1	11	1110	kkkk		C, DC, Z	
ANDLW	k	AND literal with W	1	11		kkkk		Z	
IORLW	k	Inclusive OR literal with W	1	11		kkkk		Z	
MOVLB	k	Move literal to BSR	1	00	0000	001k			
MOVLP	k	Move literal to PCLATH	1	11		1kkk			
MOVLW	k	Move literal to W	1	11		kkkk			
SUBLW	k	Subtract W from literal	1	11	1100	kkkk		C, DC, Z	
XORLW	k	Exclusive OR literal with W	1	11	1010	kkkk	kkkk	Z	

TABLE 35-3: INSTRUCTION SET

Note 1: If the Program Counter (PC) is modified, or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

2: If this instruction addresses an INDF register and the MSb of the corresponding FSR is set, this instruction will require one additional instruction cycle.

Mnemonic, Operands		nic, Description		14-Bi		t Opcode		Status	
		Description	Cycles	MSb			LSb	Affected	Notes
		CONTROL OPER	ATIONS						
BRA	k	Relative Branch	2	11	001k	kkkk	kkkk		
BRW	-	Relative Branch with W	2	00	0000	0000	1011		
CALL	k	Call Subroutine	2	10	0kkk	kkkk	kkkk		
CALLW	-	Call Subroutine with W	2	00	0000	0000	1010		
GOTO	k	Go to address	2	10	1kkk	kkkk	kkkk		
RETFIE	k	Return from interrupt	2	00	0000	0000	1001		
RETLW	k	Return with literal in W	2	11	0100	kkkk	kkkk		
RETURN	-	Return from Subroutine	2	00	0000	0000	1000		
		INHERENT OPER	ATIONS						
CLRWDT	_	Clear Watchdog Timer	1	00	0000	0110	0100	TO, PD	
NOP	-	No Operation	1	00	0000	0000	0000		
OPTION	_	Load OPTION_REG register with W	1	00	0000	0110	0010		
RESET	-	Software device Reset	1	00	0000	0000	0001		
SLEEP	-	Go into Standby mode	1	00	0000	0110	0011	TO, PD	
TRIS	f	Load TRIS register with W	1	00	0000	0110	Offf		
		C-COMPILER OPT	IMIZED						
ADDFSR	n, k	Add Literal k to FSRn	1	11	0001	0nkk	kkkk		
MOVIW	n mm	Move Indirect FSRn to W with pre/post inc/dec	1	00	0000	0001	0nmm	Z	2, 3
		modifier, mm							
	k[n]	Move INDFn to W, Indexed Indirect.	1	11	1111	0nkk	kkkk	Z	2
MOVWI	n mm	Move W to Indirect FSRn with pre/post inc/dec	1	00	0000	0001	lnmm		2, 3
		modifier, mm							
	k[n]	Move W to INDFn, Indexed Indirect.	1	11	1111	1nkk	kkkk		2

TABLE 35-3: INSTRUCTION SET (CONTINUED)

Note 1: If the Program Counter (PC) is modified, or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

2: If this instruction addresses an INDF register and the MSb of the corresponding FSR is set, this instruction will require one additional instruction cycle.

3: See Table in the MOVIW and MOVWI instruction descriptions.

35.2 Instruction Descriptions

ADDFSR	Add Literal to FSRn
Syntax:	[label] ADDFSR FSRn, k
Operands:	-32 ≤ k ≤ 31 n ∈ [0, 1]
Operation:	$FSR(n) + k \rightarrow FSR(n)$
Status Affected:	None
Description:	The signed 6-bit literal 'k' is added to the contents of the FSRnH:FSRnL register pair.
	FSRn is limited to the range 0000h-FFFFh. Moving beyond these bounds will cause the FSR to

ANDLW	AND literal with W
Syntax:	[<i>label</i>] ANDLW k
Operands:	$0 \leq k \leq 255$
Operation:	(W) .AND. (k) \rightarrow (W)
Status Affected:	Z
Description:	The contents of W register are AND'ed with the 8-bit literal 'k'. The result is placed in the W register.

ADDLW	Add literal and W
Syntax:	[<i>label</i>] ADDLW k
Operands:	$0 \leq k \leq 255$
Operation:	$(W) + k \to (W)$
Status Affected:	C, DC, Z
Description:	The contents of the W register are added to the 8-bit literal 'k' and the result is placed in the W register.

wrap-around.

ANDWF	AND W with f
Syntax:	[<i>label</i>] ANDWF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(W) .AND. (f) \rightarrow (destination)
Status Affected:	Z
Description:	AND the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

ADDWF	Add W and f
Syntax:	[<i>label</i>] ADDWF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(W) + (f) \rightarrow (destination)
Status Affected:	C, DC, Z
Description:	Add the contents of the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

ASRF	Arithmetic Right Shift		
Syntax:	[label]ASRF f{,d}		
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$		
Operation:	(f<7>)→ dest<7> (f<7:1>) → dest<6:0>, (f<0>) → C,		
Status Affected:	C, Z		
Description:	The contents of register 'f' are shifted one bit to the right through the Carry flag. The MSb remains unchanged. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is stored back in register 'f'.		



ADDWFC ADD W and CARRY bit to	f
-------------------------------	---

Syntax:	[<i>label</i>] ADDWFC f {,d}
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	$(W) + (f) + (C) \rightarrow dest$
Status Affected:	C, DC, Z
Description:	Add W, the Carry flag and data mem- ory location 'f'. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed in data memory location 'f'.

BCF	Bit Clear f
Syntax:	[<i>label</i>] BCF f,b
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b \leq 7 \end{array}$
Operation:	$0 \rightarrow (f \le b >)$
Status Affected:	None
Description:	Bit 'b' in register 'f' is cleared.

BTFSC	Bit Test f, Skip if Clear		
Syntax:	[label]BTFSC f,b		
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b \leq 7 \end{array}$		
Operation:	skip if (f) = 0		
Status Affected:	None		
Description:	If bit 'b' in register 'f' is '1', the next instruction is executed. If bit 'b', in register 'f', is '0', the next instruction is discarded, and a NOP is executed instead, making this a 2-cycle instruction.		

BRA	Relative Branch
Syntax:	[<i>label</i>]BRA label [<i>label</i>]BRA \$+k
Operands:	-256 \leq label - PC + 1 \leq 255 -256 \leq k \leq 255
Operation:	$(PC) + 1 + k \rightarrow PC$
Status Affected:	None
Description:	Add the signed 9-bit literal 'k' to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 1 + k. This instruction is a 2-cycle instruction. This branch has a limited range.

BTFSS	Bit Test f, Skip if Set
Syntax:	[label] BTFSS f,b
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b < 7 \end{array}$
Operation:	skip if (f) = 1
Status Affected:	None
Description:	If bit 'b' in register 'f' is '0', the next instruction is executed. If bit 'b' is '1', then the next instruction is discarded and a NOP is executed instead, making this a 2-cycle instruction.

Syntax:	[label] BRW
Operands:	None
Operation:	$(PC) + (W) \to PC$
Status Affected:	None
Description:	Add the contents of W (unsigned) to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 1 + (W). This instruction is a 2-cycle instruction.

BSF	Bit Set f
Syntax:	[label]BSF f,b
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b \leq 7 \end{array}$
Operation:	$1 \rightarrow (f \le b >)$
Status Affected:	None
Description:	Bit 'b' in register 'f' is set.

CALL	Call Subroutine
Syntax:	[<i>label</i>] CALL k
Operands:	$0 \leq k \leq 2047$
Operation:	(PC)+ 1→ TOS, k → PC<10:0>, (PCLATH<6:3>) → PC<14:11>
Status Affected:	None
Description:	Call Subroutine. First, return address (PC + 1) is pushed onto the stack. The 11-bit immediate address is loaded into PC bits <10:0>. The upper bits of the PC are loaded from PCLATH. CALL is a 2-cycle instruction.

CLRWDT	Clear Watchdog Timer
Syntax:	[label] CLRWDT
Operands:	None
Operation:	$\begin{array}{l} \text{O0h} \rightarrow \text{WDT} \\ \text{0} \rightarrow \underline{\text{WDT}} \text{ prescaler}, \\ \text{1} \rightarrow \underline{\overline{\text{TO}}} \\ \text{1} \rightarrow \overline{\text{PD}} \end{array}$
Status Affected:	TO, PD
Description:	CLRWDT instruction resets the Watch- dog Timer. It also resets the prescaler of the WDT. Status bits \overline{TO} and PD are set.

CALLW	Subroutine Call With W
Syntax:	[label] CALLW
Operands:	None
Operation:	(PC) +1 \rightarrow TOS, (W) \rightarrow PC<7:0>, (PCLATH<6:0>) \rightarrow PC<14:8>
Status Affected:	None
Description:	Subroutine call with W. First, the return address (PC + 1) is pushed onto the return stack. Then, the contents of W is loaded into PC<7:0>, and the contents of PCLATH into PC<14:8>. CALLW is a 2-cycle instruction.

COMF	Complement f
Syntax:	[<i>label</i>] COMF f,d
Operands:	$\begin{array}{l} 0\leq f\leq 127\\ d\in [0,1] \end{array}$
Operation:	$(\overline{f}) \rightarrow (destination)$
Status Affected:	Z
Description:	The contents of register 'f' are complemented. If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'.

CLRF	Clear f
Syntax:	[label] CLRF f
Operands:	$0 \leq f \leq 127$
Operation:	$\begin{array}{l} 00h \rightarrow (f) \\ 1 \rightarrow Z \end{array}$
Status Affected:	Z
Description:	The contents of register 'f' are cleared and the Z bit is set.

DECF	Decrement f
Syntax:	[label] DECF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(f) - 1 \rightarrow (destination)
Status Affected:	Z
Description:	Decrement register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

CLRW	Clear W
Syntax:	[label] CLRW
Operands:	None
Operation:	$\begin{array}{l} \text{O0h} \rightarrow (\text{W}) \\ 1 \rightarrow \text{Z} \end{array}$
Status Affected:	Z
Description:	W register is cleared. Zero bit (Z) is set.

DECFSZ	Decrement f, Skip if 0
Syntax:	[label] DECFSZ f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(f) - 1 \rightarrow (destination); skip if result = 0
Status Affected:	None
Description:	The contents of register 'f' are decre- mented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'. If the result is '1', the next instruction is executed. If the result is '0', then a NOP is executed instead, making it a 2-cycle instruction.

GOTO	Unconditional Branch
Syntax:	[<i>label</i>] GOTO k
Operands:	$0 \leq k \leq 2047$
Operation:	$k \rightarrow PC < 10:0>$ PCLATH<6:3> \rightarrow PC<14:11>
Status Affected:	None
Description:	GOTO is an unconditional branch. The 11-bit immediate value is loaded into PC bits <10:0>. The upper bits of PC are loaded from PCLATH<4:3>. GOTO is a 2-cycle instruction.

INCFSZ	Increment f, Skip if 0
Syntax:	[label] INCFSZ f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(f) + 1 \rightarrow (destination), skip if result = 0
Status Affected:	None
Description:	The contents of register 'f' are incre- mented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'. If the result is '1', the next instruction is executed. If the result is '0', a NOP is executed instead, making it a 2-cycle instruction.

IORLW	Inclusive OR literal with W
Syntax:	[<i>label</i>] IORLW k
Operands:	$0 \leq k \leq 255$
Operation:	(W) .OR. $k \rightarrow$ (W)
Status Affected:	Z
Description:	The contents of the W register are OR'ed with the 8-bit literal 'k'. The result is placed in the W register.

INCF	Increment f
Syntax:	[label] INCF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(f) + 1 \rightarrow (destination)
Status Affected:	Z
Description:	The contents of register 'f' are incre- mented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.

IORWF	Inclusive OR W with f
Syntax:	[<i>label</i>] IORWF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(W) .OR. (f) \rightarrow (destination)
Status Affected:	Z
Description:	Inclusive OR the W register with regis- ter 'f'. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.

LSLF	Logical Left Shift	MOVF	Move f
Syntax:	[<i>label</i>]LSLF f{,d}	Syntax:	[<i>label</i>] MOVF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$	Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	$(f < 7 >) \rightarrow C$	Operation:	$(f) \rightarrow (dest)$
	(f<6:0>) → dest<7:1> 0 → dest<0>	Status Affected:	Z
Status Affected: Description:	C, Z The contents of register 'f' are shifted one bit to the left through the Carry flag. A '0' is shifted into the LSb. If 'd' is '0', the result is placed in W. If 'd' is '1', the	Description:	The contents of register f is moved to a destination dependent upon the status of d. If $d = 0$, destination is W register. If $d = 1$, the destination is file register f itself. $d = 1$ is useful to test a file register since status flag Z is
	result is stored back in register 'f'.	Words:	affected.
	C register f		
		Cycles:	1
		Example:	MOVF FSR, 0
LSRF	Logical Right Shift		After Instruction W = value in FSR register
Syntax:	[<i>label</i>]LSRF f{,d}		Z = 1

	d ∈ [0,1]
Operation:	0 → dest<7> (f<7:1>) → dest<6:0>, (f<0>) → C,
Status Affected:	C, Z
Description:	The contents of register 'f' are shifted one bit to the right through the Carry flag. A '0' is shifted into the MSb. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is stored back in register 'f'.

 $0 \leq f \leq 127$

Operands:

► C 0→ register f

MOVIW	Move INDFn to W
Syntax:	[<i>label</i>] MOVIW ++FSRn [<i>label</i>] MOVIWFSRn [<i>label</i>] MOVIW FSRn++ [<i>label</i>] MOVIW FSRn [<i>label</i>] MOVIW k[FSRn]
Operands:	n ∈ [0,1] mm ∈ [00,01, 10, 11] -32 ≤ k ≤ 31
Operation:	$\begin{split} &\text{INDFn} \rightarrow W \\ &\text{Effective address is determined by} \\ &\text{•} \ &\text{FSR + 1 (preincrement)} \\ &\text{•} \ &\text{FSR - 1 (predecrement)} \\ &\text{•} \ &\text{FSR + k (relative offset)} \\ &\text{After the Move, the FSR value will be} \\ &\text{either:} \\ &\text{•} \ &\text{FSR + 1 (all increments)} \\ &\text{•} \ &\text{FSR - 1 (all decrements)} \\ &\text{•} \ &\text{Unchanged} \end{split}$
Status Affected:	Z

Mode	Syntax	mm
Preincrement	++FSRn	00
Predecrement	FSRn	01
Postincrement	FSRn++	10
Postdecrement	FSRn	11

Description:

This instruction is used to move data between W and one of the indirect registers (INDFn). Before/after this move, the pointer (FSRn) is updated by pre/post incrementing/decrementing it.

Note: The INDFn registers are not physical registers. Any instruction that accesses an INDFn register actually accesses the register at the address specified by the FSRn.

FSRn is limited to the range 0000h -FFFFh. Incrementing/decrementing it beyond these bounds will cause it to wrap-around.

Syntax:	[<i>label</i>] MOVLB k
Operands:	$0 \leq k \leq 31$
Operation:	$k \rightarrow BSR$
Status Affected:	None
Description:	The 5-bit literal 'k' is loaded into the Bank Select Register (BSR).

MOVLP	Move literal to PCLATH	
Syntax:	[<i>label</i>] MOVLP k	
Operands:	$0 \le k \le 127$	
Operation:	$k \rightarrow PCLATH$	
Status Affected:	None	
Description:	The 7-bit literal 'k' is loaded into the PCLATH register.	
MOVLW	Move literal to W	
Syntax:	[<i>label</i>] MOVLW k	
Operands:	$0 \le k \le 255$	
Operation:	$k \rightarrow (W)$	
Status Affected:	None	
Description:	The 8-bit literal 'k' is loaded into W reg- ister. The "don't cares" will assemble as '0's.	
Words:	1	
Cycles:	1	
Example:	MOVLW 0x5A	
	After Instruction W = 0x5A	
MOVWF	Move W to f	
Syntax:	[<i>label</i>] MOVWF f	
Operands:	$0 \leq f \leq 127$	
Operation:	$(W) \rightarrow (f)$	
Status Affected:	None	
Description:	Move data from W register to register 'f'.	
Words:	1	
Cycles:	1	
Example:	MOVWF OPTION_REG	
	Before Instruction OPTION_REG = 0xFF W = 0x4F	
	After Instruction	

After Instruction OPTION_REG = 0x4F W = 0x4F

ΜΟΥΨΙ	Move W to INDFn	
Syntax:	[<i>label</i>] MOVWI ++FSRn [<i>label</i>] MOVWIFSRn [<i>label</i>] MOVWI FSRn++ [<i>label</i>] MOVWI FSRn [<i>label</i>] MOVWI k[FSRn]	
Operands:	$\begin{array}{l} n \in [0,1] \\ mm \in [00,01,10,11] \\ -32 \leq k \leq 31 \end{array}$	
Operation:	$\label{eq:states} \begin{array}{l} W \rightarrow INDFn \\ Effective \ address \ is \ determined \ by \\ \bullet \ FSR + 1 \ (preincrement) \\ \bullet \ FSR + 1 \ (predecrement) \\ \bullet \ FSR + k \ (relative \ offset) \\ After \ the \ Move, \ the \ FSR \ value \ will \ be \\ either: \\ \bullet \ FSR + 1 \ (all \ increments) \\ \bullet \ FSR + 1 \ (all \ increments) \\ \bullet \ FSR - 1 \ (all \ decrements) \\ Unchanged \end{array}$	

Status Affected:

Mode	Syntax	mm
Preincrement	++FSRn	00
Predecrement	FSRn	01
Postincrement	FSRn++	10
Postdecrement	FSRn	11

None

Description:

This instruction is used to move data between W and one of the indirect registers (INDFn). Before/after this move, the pointer (FSRn) is updated by pre/post incrementing/decrementing it.

Note: The INDFn registers are not physical registers. Any instruction that accesses an INDFn register actually accesses the register at the address specified by the FSRn.

FSRn is limited to the range 0000h-FFFFh. Incrementing/decrementing it beyond these bounds will cause it to wrap-around.

The increment/decrement operation on FSRn WILL NOT affect any Status bits.

NOP	No Operation	
Syntax:	[<i>label</i>] NOP	
Operands:	None	
Operation:	No operation	
Status Affected:	None	
Description:	No operation.	
Words:	1	
Cycles:	1	
Example:	NOP	

OPTION	Load OPTION_REG Register with W
Syntax:	[label] OPTION
Operands:	None
Operation:	$(W) \rightarrow OPTION_REG$
Status Affected:	None
Description:	Move data from W register to OPTION_REG register.
Words:	1
Cycles:	1
Example:	OPTION
	Before Instruction OPTION_REG = 0xFF W = 0x4F After Instruction OPTION_REG = 0x4F W = 0x4F

RESET	Software Reset	
Syntax:	[label] RESET	
Operands:	None	
Operation:	Execute a device Reset. Resets the \overline{RI} flag of the PCON register.	
Status Affected:	None	
Description:	This instruction provides a way to execute a hardware Reset by software.	

RETFIE	Return from Interrupt		
Syntax:	[<i>label</i>] RETFIE k		
Operands:	None		
Operation:	$\begin{array}{l} TOS \to PC, \\ 1 \to GIE \end{array}$		
Status Affected:	None		
Description:	Return from Interrupt. Stack is POPed and Top-of-Stack (TOS) is loaded in the PC. Interrupts are enabled by setting Global Interrupt Enable bit, GIE (INTCON<7>). This is a 2-cycle instruction.		
Words:	1		
Cycles:	2		
Example:	RETFIE		
	After Interrupt PC = TOS GIE = 1		

RETURN	Return from Subroutine		
Syntax:	[label] RETURN		
Operands:	None		
Operation:	$TOS \rightarrow PC$		
Status Affected:	None		
Description:	Return from subroutine. The stack is POPed and the top of the stack (TOS) is loaded into the program counter. This is a 2-cycle instruction.		

RETLW	Return with literal in W		Detete Left fithmenuch Opmen	
Syntax:	[<i>label</i>] RETLW k	RLF	Rotate Left f through Carry	
Operands:	$0 \le k \le 255$	Syntax:	[<i>label</i>] RLF f,d	
Operation:	$k \rightarrow (W);$ TOS \rightarrow PC	Operands:	$0 \le f \le 127$ $d \in [0,1]$	
Status Affected:	None	Operation:	See description below	
Description:		Status Affected:	С	
Description.	: The W register is loaded with the 8-bit literal 'k'. The program counter is loaded from the top of the stack (the return address). This is a 2-cycle instruction.	Description:	The contents of register 'f' are rotated one bit to the left through the Carry flag. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is	
Words:	1		stored back in register 'f'.	
Cycles:	2			
Example:	CALL TABLE;W contains table	Words:	1	
	;offset value • ;W now has table value	Cycles:	1	
TABLE	•	Example:	RLF REG1,0	
	• ADDWF PC ;W = offset RETLW k1 ;Begin table RETLW k2 ; • • • RETLW kn ; End of table		Before Instruction REG1 = 1110 0110 C = 0 After Instruction REG1 = 1110 0110 W = 1100 1100 C = 1 1100	
	Before Instruction W = 0x07 After Instruction W = value of k8			

RRF	Rotate Right f through Carry		
Syntax:	[<i>label</i>] RRF f,d		
Operands:	$0 \le f \le 127$ $d \in [0,1]$		
Operation:	See description below		
Status Affected:	С		
Description:	C The contents of register 'f' are rotated one bit to the right through the Carry flag. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.		



SUBLW	Subtract W from literal			
Syntax:	[label] St	JBLW k		
Operands:	$0 \leq k \leq 255$	$0 \le k \le 255$		
Operation:	$k - (W) \to (W$	$k - (W) \to (W)$		
Status Affected:	C, DC, Z			
Description:	The W register is subtracted (2's complement method) from the 8-bit literal 'k'. The result is placed in the W register.			
	C = 0	W > k		
	C = 1	$W \leq k$		
	DC = 0	W<3:0> > k<3:0>		

DC = 1

 $W<3:0> \le k<3:0>$

SLEEP	Enter Sleep mode	
Syntax:	[label] SLEEP	
Operands:	None	
Operation:	$\begin{array}{l} \text{00h} \rightarrow \text{WDT,} \\ 0 \rightarrow \underline{\text{WDT}} \text{ prescaler,} \\ 1 \rightarrow \underline{\text{TO}}, \\ 0 \rightarrow \overline{\text{PD}} \end{array}$	
Status Affected:	TO, PD	
Description:	TO, PD The power-down Status bit, <u>PD</u> is cleared. Time-out Status bit, <u>TO</u> is set. Watchdog Timer and its prescaler are cleared. The processor is put into Sleep mode with the oscillator stopped.	

SUBWF	Subtract W from f		
Syntax:	[label] SL	IBWF f,d	
Operands:	$0 \le f \le 127$ $d \in [0,1]$		
Operation:	(f) - (W) \rightarrow (d	estination)	
Status Affected:	C, DC, Z		
Description:	Subtract (2's complement method) W register from register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f.		
	C = 0	W > f	
	C = 1	$W \leq f$	
	DC = 0	W<3:0> > f<3:0>	
	DC = 1	$W<3:0> \le f<3:0>$	

SUBWFB	Subtract W from f with Borrow
Syntax:	SUBWFB f {,d}
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	$(f) - (W) - (\overline{B}) \rightarrow dest$
Status Affected:	C, DC, Z
Description:	Subtract W and the BORROW flag (CARRY) from register 'f' (2's complement method). If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'.

SWAPF	Swap Nibbles in f
Syntax:	[label] SWAPF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	$(f<3:0>) \rightarrow (destination<7:4>),$ $(f<7:4>) \rightarrow (destination<3:0>)$
Status Affected:	None
Description:	The upper and lower nibbles of register 'f' are exchanged. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed in register 'f'.

XORLW	Exclusive OR literal with W	
Syntax:	[label] XORLW k	
Operands:	$0 \leq k \leq 255$	
Operation:	(W) .XOR. $k \rightarrow$ (W)	
Status Affected:	Z	
Description:	The contents of the W register are XOR'ed with the 8-bit literal 'k'. The result is placed in the W register.	

TRIS	Load TRIS Register with W
Syntax:	[label] TRIS f
Operands:	$5 \le f \le 7$
Operation:	(W) \rightarrow TRIS register 'f'
Status Affected:	None
Description:	Move data from W register to TRIS register. When 'f' = 5, TRISA is loaded. When 'f' = 6, TRISB is loaded. When 'f' = 7, TRISC is loaded.

XORWF	Exclusive OR W with f
Syntax:	[label] XORWF f,d
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	(W) .XOR. (f) \rightarrow (destination)
Status Affected:	Z
Description:	Exclusive OR the contents of the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

36.0 ELECTRICAL SPECIFICATIONS

36.1 Absolute Maximum Ratings^(†)

Ambient temperature under bias	40°C to +125°C
Storage temperature	65°C to +150°C
Voltage on pins with respect to Vss	
on VDD pin	
PIC16F1777/8/9	0.3V to +6.5V
PIC16LF1777/8/9	0.3V to +4.0V
on MCLR pin	0.3V to +9.0V
on all other pins	-0.3V to (VDD + 0.3V)
Maximum current	
on Vss pin ⁽¹⁾	
$-40^{\circ}C \leq TA \leq +85^{\circ}C$	350 mA
-40°C \leq TA \leq +125°C	120 mA
on VDD pin ⁽¹⁾ PIC16(L)F1778 only	
$-40^{\circ}C \leq TA \leq +85^{\circ}C$	250 mA
-40°C \leq TA \leq +125°C	85 mA
on VDD pin ⁽¹⁾ PIC16(L)F1777/9 only	
$-40^{\circ}C \leq T_A \leq +85^{\circ}C$	350 mA
-40°C \leq TA \leq +125°C	120 mA
Sunk by any standard I/O pin	50 mA
Sourced by any standard I/O pin	50 mA
Sunk by any High Current I/O pin	100 mA
Sourced by any High Current I/O pin	100 mA
Sourced by any Op Amp output pin	100 mA
Clamp current, Iк (VPIN < 0 or VPIN > VDD)	±20 mA
Total power dissipation ⁽²⁾	800 mW

Note 1: Maximum current rating requires even load distribution across I/O pins. Maximum current rating may be limited by the device package power dissipation characterizations, see Table 36-6: Thermal Characteristics to calculate device specifications.

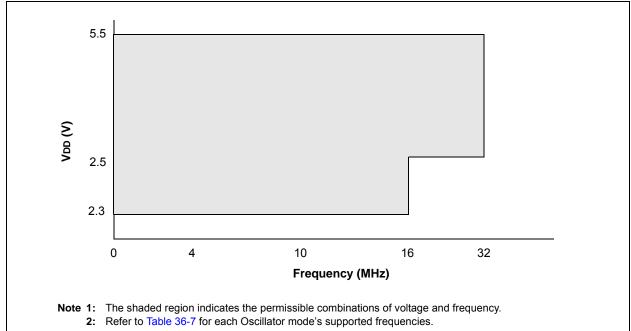
2: Power dissipation is calculated as follows: Pdis = VDD* {Idd- Σ Ioh} + Σ {VDD-Voh)*Ioh} + Σ (Vol*IoI).

† NOTICE: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure above maximum rating conditions for extended periods may affect device reliability.

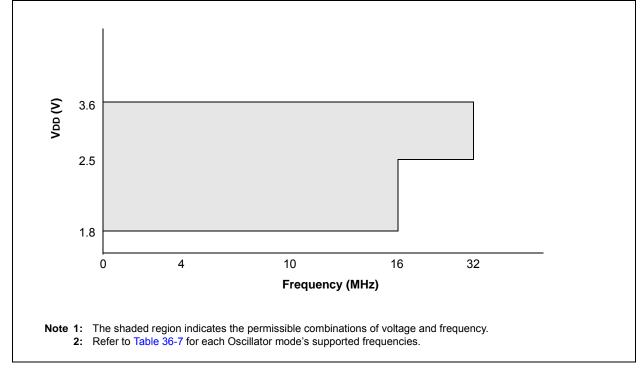
36.2 Standard Operating Conditions

The standard operating co	onditions for any device are defined as:	
Operating Voltage: Operating Temperature:	$\label{eq:VDDMAX} \begin{split} V \text{DDMIN} &\leq V \text{DD} \leq V \text{DDMAX} \\ T \text{A}_{\text{MIN}} &\leq T \text{A}_{\text{MAX}} \end{split}$	
VDD — Operating Supply	y Voltage ⁽¹⁾	
PIC16LF1777/8/9		
VDDMIN (F	Fosc \leq 16 MHz)	+1.8V
VDDMIN (F	Fosc > 16 MHz)	+2.5V
VDDMAX		+3.6V
PIC16F1777/8/9		
VDDMIN (F	Fosc \leq 16 MHz)	+2.3V
VDDMIN (F	Fosc > 16 MHz)	+2.5V
VDDMAX		+5.5V
TA — Operating Ambient	it Temperature Range	
Industrial Temperat	ture	
Та_міл		-40°C
Та_мах		+85°C
Extended Temperat	Iture	
TA_MIN		40°C
Та_мах		+125°C
Note 1: See Paramete	er D001, DS Characteristics: Supply Voltage.	









36.3 DC Characteristics

TABLE 36-1:SUPPLY VOLTAGE

PIC16LF	1777/8/9		Standard Operating Conditions (unless otherwise stated)							
PIC16F1	777/8/9									
Param. No.	Sym.	Characteristic	Min.	Тур†	Max.	Units	Conditions			
D001	Vdd	Supply Voltage		÷.						
			VDDMIN 1.8 2.5		VDDMAX 3.6 3.6	V V	Fosc ≤ 16 MHz Fosc ≤ 32 MHz			
D001		PIC16F1777/8/9	2.3 2.5	_	5.5 5.5	V V	Fosc \leq 16 MHz Fosc \leq 32 MHz			
D002*	Vdr	RAM Data Retention Voltage ⁽¹⁾								
			1.5	-	—	V	Device in Sleep mode			
D002*			1.7	_	—	V	Device in Sleep mode			
D002A*	VPOR	Power-on Reset Release Voltage								
			—	1.6	—	V				
D002A*			—	1.6	—	V				
D002B*	VPORR*	Power-on Reset Rearm Voltage ⁽²⁾								
			—	0.8	—	V				
D002B*			—	1.5	_	V				
D003	Vfvr	Fixed Voltage Reference Voltage ⁽³⁾	-4 -4 -7		+4 +4 +7	% % %	$\begin{array}{l} 1x \mbox{ gain, } 1.024, \mbox{ VDD} \geq 2.5 \mbox{V}, \mbox{ -40}^\circ \mbox{C} \ to \ +85^\circ \mbox{C} \\ 2x \mbox{ gain, } 2.048, \mbox{ VDD} \geq 2.5 \mbox{V}, \mbox{ -40}^\circ \mbox{C} \ to \ +85^\circ \mbox{C} \\ 4x \mbox{ gain, } 4.096, \mbox{ VDD} \geq 4.5 \mbox{V}, \mbox{ -40}^\circ \mbox{C} \ to \ +85^\circ \mbox{C} \end{array}$			
D004*	SVDD	VDD Rise Rate	0.05	—	—	V/ms	Ensures that the Power-on Reset signal is released properly.			

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered in Sleep mode without losing RAM data.

2: See Figure 36-3: POR and POR Rearm with Slow Rising VDD.

3: Industrial temperature range only.

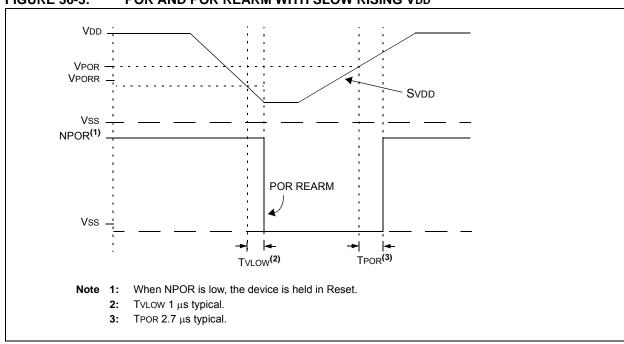


FIGURE 36-3: POR AND POR REARM WITH SLOW RISING VDD

PIC16LI	PIC16LF1777/8/9		rd Operat	ing Cond	litions (un	less oth	erwise stated)
PIC16F	1777/8/9						
Param	Device	Min.	Typ†	Max.	Units		Conditions
No.	Characteristics		iypi	max.	Units	Vdd	Note
D009	LDO Regulator	-	75	_	μA	—	High-Power mode, normal operation
		—	15		μA		Sleep, VREGCON<1> = 0
		_	0.3		μA		Sleep, VREGCON<1> = 1
D010		_	8	25	μA	1.8	Fosc = 32 kHz,
		_	12	30	μA	3.0	LP Oscillator mode (Note 4), -40°C \leq TA \leq +85°C
D010		_	21	30	μA	2.3	Fosc = 32 kHz,
		—	25	34	μA	3.0	LP Oscillator mode (Note 4,5)
		—	26	35	μA	5.0	$-40^{\circ}C \le TA \le +85^{\circ}C$
D012		_	210	440	μA	1.8	Fosc = 4 MHz,
		—	390	620	μA	3.0	XT Oscillator mode
D012			320	530	μA	2.3	Fosc = 4 MHz,
		_	430	680	μA	3.0	XT Oscillator mode (Note 5)
		—	530	790	μA	5.0	
D014		_	170	380	μA	1.8	Fosc = 4 MHz,
		—	320	550	μA	3.0	External Clock (ECM), Medium Power mode
D014		—	250	513	μA	2.3	Fosc = 4 MHz,
		—	360	645	μA	3.0	External Clock (ECM), Medium Power mode
		—	430	735	μA	5.0	
D015		—	2.5	3.8	mA	3.0	Fosc = 32 MHz,
		_	3.1	4.0	mA	3.6	External Clock (ECH), High-Power mode (Note 5)
D015		_	2.5	4.3	mA	3.0	Fosc = 32 MHz,
		-	2.7	4.6	mA	5.0	External Clock (ECH), High-Power mode (Note 5)

TABLE 36-2: SUPPLY CURRENT (IDD)^(1,2)

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The test conditions for all IDD measurements in active operation mode are: OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD; MCLR = VDD; WDT disabled.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors, such as I/O pin loading and switching rate, oscillator type, internal code execution pattern and temperature, also have an impact on the current consumption.

3: For EXTRC oscillator configurations, current through REXT is not included. The current through the resistor can be extended by the formula IR = VDD/2REXT (mA) with REXT in kΩ.

4: FVR and BOR are disabled.

5: 8 MHz crystal/oscillator with 4x PLL enabled.

TABLE 36-2: SUPPLY CURRENT (IDD)^(1,2) (CONTINUED)

PIC16LF	1777/8/9	Standa	rd Operat	ing Conc	litions (un	less oth	erwise stated)
PIC16F1	777/8/9						
Param	Device	Min.	Tunt	Max.	Units		Conditions
No.	Characteristics	WIIII.	Тур†	WidX.	Units	Vdd	Note
D017		_	115	190	μA	1.8	Fosc = 500 kHz,
		—	145	320	μA	3.0	MFINTOSC mode
D017			160	215	μA	2.3	Fosc = 500 kHz,
			180	340	μA	3.0	MFINTOSC mode
		_	230	420	μA	5.0	
D019		_	0.9	1.5	mA	1.8	Fosc = 16 MHz,
		—	1.5	2.3	mA	3.0	HFINTOSC mode
D019		_	1.2	2.0	mA	2.3	Fosc = 16 MHz,
			1.5	2.5	mA	3.0	HFINTOSC mode
		_	1.7	2.6	mA	5.0	
D020		_	2.9	4.2	mA	3.0	Fosc = 32 MHz,
		_	3.5	4.3	mA	3.6	HFINTOSC mode (Note 5)
D020		_	2.9	4.2	mA	3.0	Fosc = 32 MHz,
		_	3.0	5.0	mA	5.0	HFINTOSC mode (Note 5)
D022		_	2.8	4	mA	3.0	Fosc = 32 MHz,
		_	3.4	4.7	mA	3.6	HS Oscillator mode (Note 5)
D022		_	2.9	4	mA	3.0	Fosc = 32 MHz
		—	3.1	4.5	mA	5.0	HS Oscillator mode (Note 5)

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The test conditions for all IDD measurements in active operation mode are: OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD; MCLR = VDD; WDT disabled.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors, such as I/O pin loading and switching rate, oscillator type, internal code execution pattern and temperature, also have an impact on the current consumption.

3: For EXTRC oscillator configurations, current through REXT is not included. The current through the resistor can be extended by the formula IR = VDD/2REXT (mA) with REXT in kΩ.

4: FVR and BOR are disabled.

5: 8 MHz crystal/oscillator with 4x PLL enabled.

PIC16LF1	777/8/9		Operating Conditions: (unless otherwise stated) Low-Power Sleep Mode Low-Power Sleep Mode, VREGPM = 1										
PIC16F17	77/8/9	Low-Po	ower Sle	ep Mode,	VREGPM	=1							
Param	Device Characteristics	Min.	Тур†	Max.	Max.	Units		Conditions					
No.	Device Characteristics	WIIII.	וקעי	+85°C	+125°C	Units	Vdd	Note					
D023	Base IPD		0.05	1.0	8.0	μA	1.8	WDT, BOR, FVR, and SOSC					
		—	0.08	2.0	9.0	μA	3.0	disabled, all Peripherals Inactive					
D023	Base IPD		0.3	2.4	10	μA	2.3	WDT, BOR, FVR, and SOSC					
			0.4	4	12	μA	3.0	disabled, all Peripherals Inactive, Low-Power Sleep mode					
		—	0.5	6	15	μA	5.0	Low-Power Sleep mode					
D023A	Base IPD		9.8	17	28	μA	2.3	WDT, BOR, FVR and SOSC					
		_	10.3	20	40	μA	3.0	disabled, all Peripherals inactive,					
		—	11.5	22	44	μA	5.0	Normal Power Sleep mode VREGPM = 0					
D024		_	0.5	6	14	μA	1.8	WDT Current					
		—	0.8	7	17	μA	3.0						
D024		_	0.8	6	15	μA	2.3	WDT Current					
		—	0.9	7	20	μA	3.0						
		—	1.0	8	22	μA	5.0						
D025		_	15	28	30	μA	1.8	FVR Current (ADC)					
		—	24	35	38	μA	3.0						
D025		—	18	33	35	μA	2.3	FVR Current (ADC)					
		_	24	35	40	μA	3.0						
		_	26	37	44	μA	5.0						
D025A		_	25	50	55	μA	1.8	FVR Current (DAC)					
		_	30	65	70	μA	3.0						
D025A		_	30	55	66	μA	2.3	FVR Current (DAC)					
		_	32	68	82	μA	3.0						
		_	35	77	90	μA	5.0						
D026		—	7.5	25	28	μA	3.0	BOR Current					
D026		—	10	25	28	μA	3.0	BOR Current					
		_	12	28	31	μA	5.0						
D027		_	0.5	4	10	μA	3.0	LPBOR Current					
D027		_	0.8	6	15	μA	3.0	LPBOR Current					
		—	1	8	17	μA	5.0						
D028		_	0.5	5	9	μA	1.8	SOSC Current					
			0.8	8.5	12	μA	3.0						
D028		—	1.1	6	10	μA	2.3	SOSC Current					
		—	1.3	8.5	20	μA	3.0						
			1.4	10	25	μA	5.0						

TABLE 36-3: POWER-DOWN CURRENTS (IPD)^(1,2)

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The peripheral current is the sum of the base IPD and the additional current consumed when this peripheral is enabled. The peripheral △ current can be determined by subtracting the base IDD or IPD current from this limit. Max values should be used when calculating total current consumption.

2: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to Vss.

3: ADC clock source is FRC.

PIC16LF17	777/8/9			ditions: (ເ ep Mode	unless oth	erwise s	stated)					
PIC16F177	77/8/9	Low-Po	Low-Power Sleep Mode, VREGPM = 1									
Param	Device Characteristics	Min.	Typt	Max.	Max.	Units	Conditions					
No.	Device Gilaracteristics	WIII.	пп. турт	+85°C	+125°C		Vdd	Note				
D029		-	0.05	2	9	μA	1.8	ADC Current (Note 3),				
		_	0.08	3	10	μA	3.0	no conversion in progress				
D029		—	0.3	4	12	μA	2.3	ADC Current (Note 3),				
		—	0.4	5	13	μA	3.0	no conversion in progress				
		—	0.5	7	16	μA	5.0					
D030		_	250	_	—	μA	1.8	ADC Current (Note 3),				
		—	250	—	—	μA	3.0	conversion in progress				
D030		—	280	_	—	μA	2.3	ADC Current (Note 3),				
			280	—	—	μA	3.0	conversion in progress				
		—	280		—	μA	5.0					
D031		—	250	650	—	μA	3.0	Op Amp (High power)				
D031		_	250	650	—	μA	3.0	Op Amp (High power)				
		—	350	850	—	μA	5.0					
D032		—	250	600	_	μA	1.8	Comparator, CxSP = 0				
		_	300	650	—	μA	3.0					
D032		_	280	600	—	μA		Comparator, CxSP = 0				
		_	300	650	_	μA	3.0	VREGPM = 0				
		_	310	650	_	μA	5.0					

TABLE 36-3: POWER-DOWN CURRENTS (IPD)^(1,2) (CONTINUED)

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The peripheral current is the sum of the base IPD and the additional current consumed when this peripheral is enabled. The peripheral △ current can be determined by subtracting the base IDD or IPD current from this limit. Max values should be used when calculating total current consumption.

2: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to Vss.

3: ADC clock source is FRC.

TABLE 36-4: I/O PORTS

Standard	d Operati	ing Conditions (unless otherwi	se stated)				
Param No.	Sym.	Characteristic	Min.	Тур†	Max.	Units	Conditions
	VIL	Input Low Voltage					
		I/O PORT:					
D034		with TTL buffer	_	—	0.8	V	$4.5V \leq V\text{DD} \leq 5.5V$
D034A			_	—	0.15 Vdd	V	$1.8V \leq V\text{DD} \leq 4.5V$
D035		with Schmitt Trigger buffer	_	—	0.2 Vdd	V	$2.0V \leq V\text{DD} \leq 5.5V$
		with I ² C levels	_	—	0.3 VDD	V	
		with SMBus levels	_	—	0.8	V	$2.7V \leq V\text{DD} \leq 5.5V$
D036		MCLR, OSC1 (EXTRC mode)	_	—	0.2 Vdd	V	(Note 1)
D036A		OSC1 (HS mode)	_	—	0.3 Vdd	V	
	Vih	Input High Voltage	•	•	•	•	
		I/O ports:					
D040		with TTL buffer	2.0	_	_	V	$4.5V \leq V\text{DD} \leq 5.5V$
D040A			0.25 VDD + 0.8	—	—	V	$1.8V \leq V\text{DD} \leq 4.5V$
D041		with Schmitt Trigger buffer	0.8 Vdd	—	—	V	$2.0V \leq V \text{DD} \leq 5.5 V$
		with I ² C levels	0.7 Vdd	_	_	V	
		with SMBus levels	2.1	_	_	V	$2.7V \leq V\text{DD} \leq 5.5V$
D042		MCLR	0.8 Vdd	—	—	V	
D043A		OSC1 (HS mode)	0.7 Vdd	_	—	V	
D043B		OSC1 (EXTRC oscillator)	0.9 Vdd	_	—	V	VDD > 2.0V (Note 1)
	lı∟	Input Leakage Current ⁽²⁾				•	
D060		I/O Ports	—	± 5	± 125	nA	$Vss \le VPIN \le VDD$, Pin at high-impedance, 85°C
			—	± 5	± 1000	nA	$Vss \le VPIN \le VDD$, Pin at high-impedance, 125°C
D061		MCLR ⁽³⁾	—	± 5	± 200	nA	$Vss \le VPIN \le VDD$, Pin at high-impedance, 85°C
	IPUR	Weak Pull-up Current	•	•	•	•	
D070*			25	100	200	μA	VDD = 3.3V, VPIN = VSS
	Vol	Output Low Voltage ⁽⁴⁾					
D080		Standard I/O ports	_	_	0.6	v	IOL = 8mA, VDD = 5V IOL = 6mA, VDD = 3.3V IOL = 1.8mA, VDD = 1.8V
D080A		High Drive I/O ports		_	0.6	V	Юн = 10mA, VDD = 2.3V, HIDCx = 1
		3	_	0.6	_	V	ЮН = 32mA, VDD = 3.0V, HIDCx = 1
			—	0.6	—	V	ЮН = 51mA, VDD = 5.0V, HIDCx = 1
	Voн	Output High Voltage ⁽⁴⁾					
D090		Standard I/O ports	\/ 0 7				IOH = 3.5mA, VDD = 5V
			Vdd - 0.7	—	—	V	IOH = 3mA, VDD = 3.3V IOH = 1mA, VDD = 1.8V
		High Drive I/O ports	VDD - 0.7			V	
D090A			0.7 - UUV	— Vdd - 0.7		V	ІОН = 10mA, VDD = 2.3V, HIDCx = 1 ІОН = 37mA, VDD = 3.0V, HIDCx = 1
			_	VDD - 0.7 VDD - 0.7	_	v	10H = 54MA, $VDD = 5.0V$, $HIDCx = 1$
*	These n	parameters are characterized but	not tostod	1	1	I	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: In EXTRC oscillator configuration, the OSC1/CLKIN pin is a Schmitt Trigger input. It is not recommended to use an external clock in EXTRC mode.

2: Negative current is defined as current sourced by the pin.

3: The leakage current on the MCLR pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

4: Including OSC2 in CLKOUT mode.

I/O PORTS (CONTINUED) (CONTINUED) TABLE 36-4:

Standard Operating Conditions (unless otherwise stated)									
Param No.	Sym.	Characteristic	Min.	Тур†	Max.	Units	Conditions		
Capacitive Loading Specs on Output Pins									
D101* C	COSC2	OSC2 pin	_	_	15	pF	In XT, HS and LP modes when external clock is used to drive OSC1		
D101A* C	Cio	All I/O pins			50	pF			

These parameters are characterized but not tested.

Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are t not tested.

Note 1: In EXTRC oscillator configuration, the OSC1/CLKIN pin is a Schmitt Trigger input. It is not recommended to use an external clock in EXTRC mode.

2: Negative current is defined <u>as current sourced by the pin.</u>

3: The leakage current on the MCLR pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

4: Including OSC2 in CLKOUT mode.

Standa	rd Opera	ting Conditions (unless otherw	vise stated)				
Param No.	Sym.	Characteristic	Min.	Тур†	Max.	Units	Conditions
		Program Memory Programming Specifications					
D110	Vінн	Voltage on MCLR/VPP pin	8.0	—	9.0	V	(Note 2, Note 3)
D111	IDDP	Supply Current during Programming	—	—	10	mA	
D112	VBE	VDD for Bulk Erase	2.7	—	VDDMAX	V	
D113	VPEW	VDD for Write or Row Erase	VDDMIN	_	VDDMAX	V	
D114	IPPPGM	Current on MCLR/VPP during Erase/Write		1.0	—	mA	
D115	IDDPGM	Current on VDD during Erase/Write	_	5.0	—	mA	
		Program Flash Memory					
D121	Ер	Cell Endurance	10K	—	_	E/W	-40°C ≤ TA ≤ +85°C (Note 1)
D122	VPRW	VDD for Read/Write	VDDMIN	—	VDDMAX	V	
D123	Tiw	Self-timed Write Cycle Time	—	2	2.5	ms	
D124	TRETD	Characteristic Retention	—	40	-	Year	Provided no other specifications are violated
D125	EHEFC	High-Endurance Flash Cell	100K	—	—	E/W	$-0^{\circ}C \le TA \le +60^{\circ}C$, Lower byte last 128 addresses

TABLE 36-5: MEMORY PROGRAMMING SPECIFICATIONS

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Self-write and Block Erase.

2: Required only if single-supply programming is disabled.

3: The MPLAB ICD 2 does not support variable VPP output. Circuitry to limit the ICD 2 VPP voltage must be placed between the ICD 2 and target system when programming or debugging with the ICD 2.

TABLE 36-6: THERMAL CHARACTERISTICS

Param No.	Sym.	Characteristic	Тур.	Units	Conditions
TH01	θJA	Thermal Resistance Junction to Ambient	60.0	°C/W	28-pin SPDIP package
			80.0	°C/W	28-pin SOIC package
			90.0	°C/W	28-pin SSOP package
			48	°C/W	28-pin UQFN 4x4mm package
			47.2	°C/W	40-pin PDIP package
			46.0	°C/W	44-pin TQFP package
			41.0	°C/W	40-pin UQFN 5x5mm package
TH02	θJC	Thermal Resistance Junction to Case	31.4	°C/W	28-pin SPDIP package
			24	°C/W	28-pin SOIC package
			24	°C/W	28-pin SSOP package
			12	°C/W	28-pin UQFN 4x4mm package
			24.70	°C/W	40-pin PDIP package
			14.5	°C/W	44-pin TQFP package
			5.5	°C/W	40-pin UQFN 5x5mm package
TH03	Тјмах	Maximum Junction Temperature	150	°C	
TH04	PD	Power Dissipation		W	PD = PINTERNAL + PI/O
TH05	PINTERNAL	Internal Power Dissipation	_	W	Pinternal = Idd x Vdd ⁽¹⁾
TH06	Pi/o	I/O Power Dissipation	_	W	$PI/O = \Sigma (IOL * VOL) + \Sigma (IOH * (VDD - VOH))$
TH07	Pder	Derated Power	_	W	Pder = PDmax (Τj - Τa)/θja ⁽²⁾

Standard Operating Conditions (unless otherwise stated)

Note 1: IDD is current to run the chip alone without driving any load on the output pins.

2: TA = Ambient Temperature, TJ = Junction Temperature

36.4 AC Characteristics

Timing Parameter Symbology has been created with one of the following formats:

1. TppS2ppS

2. TppS

Z. TPPS			
т			
F	Frequency	Т	Time
Lowerc	case letters (pp) and their meanings:		
рр			
сс	CCP1	osc	OSC1
ck	CLKOUT	rd	RD
CS	CS	rw	RD or WR
di	SDI	sc	SCK
do	SDO	SS	SS
dt	Data in	tO	TOCKI
io	I/O PORT	t1	T1CKI
mc	MCLR	wr	WR
Upperc	case letters and their meanings:		
S			
F	Fall	Р	Period
н	High	R	Rise
I	Invalid (High-impedance)	V	Valid
L	Low	Z	High-impedance

FIGURE 36-4: LOAD CONDITIONS

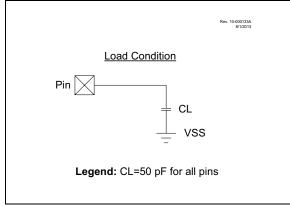


FIGURE 36-5: CLOCK TIMING

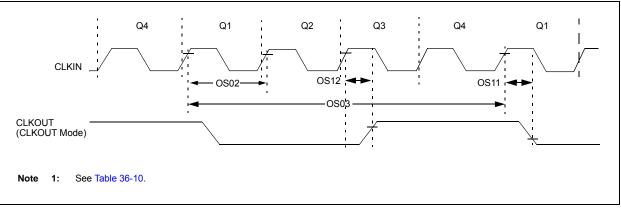


TABLE 36-7: CLOCK OSCILLATOR TIMING REQUIREMENTS

Param No.	Sym.	Characteristic	Min.	Тур†	Max.	Units	Conditions
OS01	Fosc	External CLKIN Frequency ⁽¹⁾	DC	—	0.5	MHz	External Clock (ECL)
			DC	—	4	MHz	External Clock (ECM)
			DC	—	32	MHz	External Clock (ECH)
		Oscillator Frequency ⁽¹⁾	_	32.768	-	kHz	LP Oscillator
			0.1	—	4	MHz	XT Oscillator
			1	—	4	MHz	HS Oscillator
			1	—	20	MHz	HS Oscillator, VDD > 2.7V
			DC	—	4	MHz	EXTRC, VDD > 2.0V
OS02	Tosc	External CLKIN Period ⁽¹⁾	27	—	×	μS	LP Oscillator
			250	—	∞	ns	XT Oscillator
			50	—	∞	ns	HS Oscillator
			50	—	∞	ns	External Clock (EC)
		Oscillator Period ⁽¹⁾	_	30.5	_	μS	LP Oscillator
			250	—	10,000	ns	XT Oscillator
			50	—	1,000	ns	HS Oscillator
			31.25	—	—	ns	EXTRC
OS03	Тсү	Instruction Cycle Time ⁽¹⁾	125	TCY	DC	ns	Tcy = 4/Fosc
OS04*	TosH,	External CLKIN High,	2		_	μS	LP Oscillator
	TosL	External CLKIN Low	100	—	—	ns	XT Oscillator
			20	—	—	ns	HS Oscillator
OS05*	TosR,	External CLKIN Rise,	0	—	∞	ns	LP Oscillator
	TosF	External CLKIN Fall	0	—	∞	ns	XT Oscillator
			0	—	∞	ns	HS Oscillator

Standard Operating Conditions (unless otherwise stated)

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Instruction cycle period (TCY) equals four times the input oscillator time base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min" values with an external clock applied to OSC1 pin. When an external clock input is used, the "max" cycle time limit is "DC" (no clock) for all devices.

TABLE 36-8: OSC	LATOR PARAMETERS
-----------------	------------------

Standar	Standard Operating Conditions (unless otherwise stated)									
Param No.	Sym.	Characteristic	Freq. Tolerance	Min.	Тур†	Max.	Units	Conditions		
OS08	HFosc	Internal Calibrated HFINTOSC Frequency ⁽¹⁾	±2%		16.0	—	MHz	VDD = 3.0V, TA = 25°C, (Note 2)		
OS08A	MFosc	Internal Calibrated MFINTOSC Frequency ⁽¹⁾	±2%	_	500	—	kHz	VDD = 3.0V, TA = 25°C, (Note 2)		
OS09	LFosc	Internal LFINTOSC Frequency	_	_	31	_	kHz	$-40^\circ C \le TA \le +125^\circ C$		
OS10*	TWARM	HFINTOSC Wake-up from Sleep Start-up Time	_	_	3.2	8	μS			
		MFINTOSC Wake-up from Sleep Start-up Time	—	—	24	35	μS			
	TLFOSC ST	LFINTOSC Wake-up from Sleep Start-up Time	—	_	0.5	_	ms			

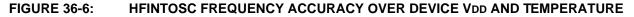
* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: To ensure these oscillator frequency tolerances, VDD and VSS must be capacitively decoupled as close to the device as possible. 0.1 μ F and 0.01 μ F values in parallel are recommended.

2: See Figure 36-6: HFINTOSC Frequency Accuracy Over Device VDD and Temperature, Figure 37-75: Wake From Sleep, VREGPM = 0., and Figure 36-6: HFINTOSC Frequency Accuracy Over Device VDD and Temperature.

3: See Figure 37-58: LFINTOSC Frequency, PIC16LF1777/8/9 Only., and Figure 37-59: LFINTOSC Frequency, PIC16F1777/8/9 Only..



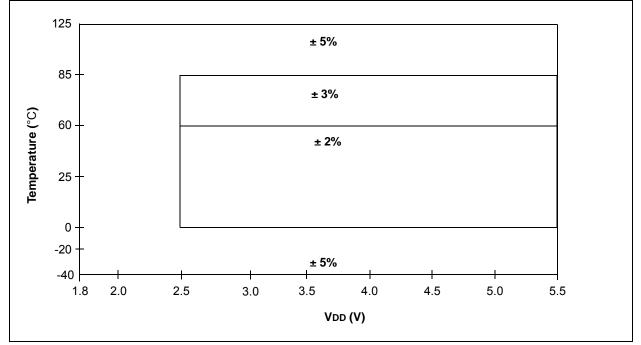
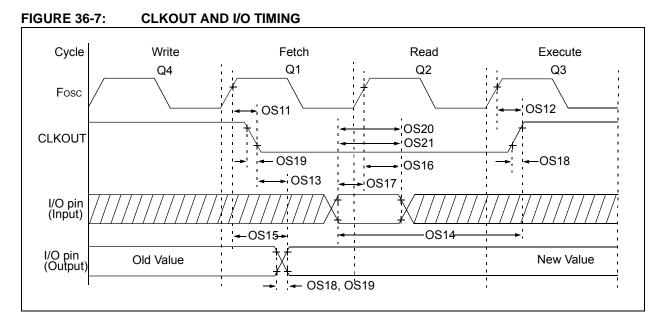


TABLE 36-9: PLL CLOCK TIMING SPECIFICATIONS

Standar	Standard Operating Conditions (unless otherwise stated)									
Param No.	Sym.	Characteristic	Min.	Тур†	Max.	Units	Conditions			
F10	Fosc	Oscillator Frequency Range	4	_	8	MHz				
F11	Fsys	On-Chip VCO System Frequency	16	_	32	MHz				
F12	TRC	PLL Start-up Time (Lock Time)		_	2	ms				
F13*	ΔCLK	CLKOUT Stability (Jitter)	-0.25%		+0.25%	%				
*	Thes	e parameters are characterized but not tested	_							

These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.



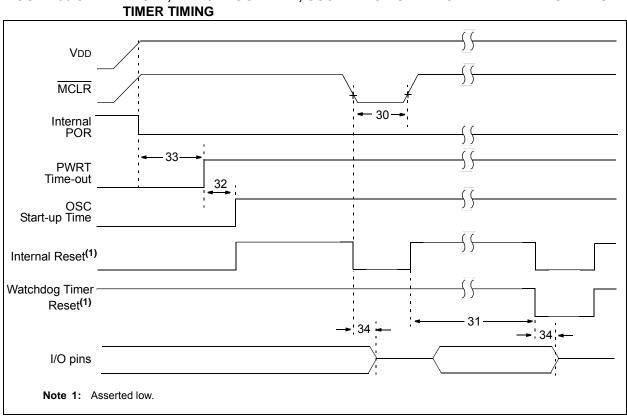
Standar	d Operating	Conditions (unless otherwise stated)						
Param No.	Sym.	Characteristic	Min.	Тур†	Max.	Units	Conditions	
OS11	TosH2ckL	Fosc↑ to CLKOUT↓ ⁽¹⁾	—	_	70	ns	$3.3V \le V\text{DD} \le 5.0V$	
OS12	TosH2ckH	Fosc↑ to CLKOUT↑ ⁽¹⁾	—	_	72	ns	$3.3V \le V\text{DD} \le 5.0V$	
OS13	TckL2ioV	CLKOUT↓ to Port out valid ⁽¹⁾	—	_	20	ns		
OS14	TioV2ckH	Port input valid before CLKOUT ⁽¹⁾	Tosc + 200 ns			ns		
OS15	TosH2ioV	Fosc↑ (Q1 cycle) to Port out valid	—	50	70*	ns	$3.3V \le V\text{DD} \le 5.0V$	
OS16	TosH2iol	Fosc↑ (Q2 cycle) to Port input invalid (I/O in hold time)	50	—	—	ns	$3.3V \le V\text{DD} \le 5.0V$	
OS17	TioV2osH	Port input valid to Fosc↑ (Q2 cycle) (I/O in setup time)	20	—	—	ns		
OS18*	TioR	Port output rise time ⁽²⁾		40 15	72 32	ns	$\begin{array}{l} VDD = 1.8V \\ 3.3V \leq VDD \leq 5.0V \end{array}$	
OS19*	TioF	Port output fall time ⁽²⁾		28 15	55 30	ns	$\begin{array}{l} VDD \mbox{=} 1.8V \\ 3.3V \le VDD \le 5.0V \end{array}$	
OS20*	Tinp	INT pin input high or low time	25			ns		
OS21*	Tioc	Interrupt-on-change new input level time	25		_	ns		

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated.

Note 1: Measurements are taken in EXTRC mode where CLKOUT output is 4 x Tosc.

2: Slew rate limited.



RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND POWER-UP **FIGURE 36-8:**

TABLE 36-11: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER, POWER-UP TIMER AND BROWN-OUT RESET PARAMETERS

Standa	rd Opera	ting Conditions (unless otherwise s	tated)				
Param No.	Sym.	Characteristic	Min.	Тур†	Max.	Units	Conditions
30	ТмсL	MCLR Pulse Width (low)	2			μS	
31	TWDTLP	Low-Power Watchdog Timer Time-out Period	10	16	27	ms	VDD = 3.3V-5V 1:512 Prescaler used
32	Tost	Oscillator Start-up Timer Period ⁽¹⁾		1024		Tosc	
33*	TPWRT	Power-up Timer Period, $\overline{PWRTE} = 0$	40	65	140	ms	
34*	Tioz	I/O high-impedance from MCLR Low or Watchdog Timer Reset	_	—	2.0	μS	
35	VBOR	Brown-out Reset Voltage ⁽²⁾	2.55	2.70	2.85	V	BORV = 0
			2.30 1.80	2.45 1.90	2.60 2.10	V V	BORV = 1 (PIC16F1777/8/9) BORV = 1 (PIC16LF1777/8/9)
35A	VLPBOR	Low-Power Brown-out	1.8	2.1	2.5	V	LPBOR = 1
36*	VHYST	Brown-out Reset Hysteresis	0	25	75	mV	$-40^\circ C \le TA \le +85^\circ C$
37*	TBORDC	Brown-out Reset DC Response Time	1	3	35	μS	$VDD \leq VBOR$

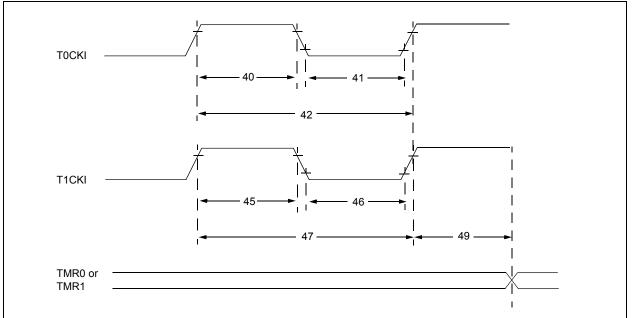
* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: By design, the Oscillator Start-up Timer (OST) counts the first 1024 cycles, independent of frequency.

2: To ensure these voltage tolerances, VDD and Vss must be capacitively decoupled as close to the device as possible. 0.1 μ F and 0.01 μ F values in parallel are recommended.

FIGURE 36-9: TIMER0 AND TIMER1 EXTERNAL CLOCK TIMINGS





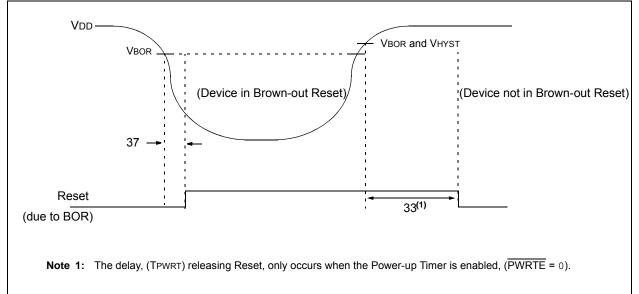


TABLE 36-12: TIMER0 AND TIMER1 EXTERNAL CLOCK REQUIREMENTS

	r d Operating ng Temperatur	•	inless otherwis ≤ +125°C	e stated)					
Param No.	Sym.		Characteristic		Min.	Тур†	Max.	Units	Conditions
40*	TT0H T0CKI High Pulse Width No Prescaler		0.5 Tcy + 20	_	_	ns			
				With Prescaler	10	—		ns	
41*	TT0L	T0CKI Low F	ulse Width	No Prescaler	0.5 Tcy + 20	—		ns	
		With Prescaler		10	—		ns		
42*	Тт0Р	T0CKI Period			Greater of: 20 or <u>Tcy + 40</u> N	—	—	ns	N = prescale value
45*	T⊤1H	T1CKI High	Synchronous, No Prescaler		0.5 Tcy + 20	—	_	ns	
		Time	Synchronous, with Prescaler		15	_		ns	
			Asynchronous		30	_		ns	
46*	TT1L	T1CKI Low Time	Synchronous, No Prescaler		0.5 Tcy + 20	_		ns	
			Synchronous, with Prescaler		15	—	—	ns	
			Asynchronous		30	_		ns	
47*	TT1P	T1CKI Input Period	Synchronous		Greater of: 30 or (Tcy + 40)*N	—	_	ns	N = prescale value
			Asynchronous		60	_		ns	
48	F⊤1	,	scillator Input Freadback		32.4	32.768	33.1	kHz	
49*	TCKEZTMR1	Delay from E Increment	xternal Clock Ec	lge to Timer	2 Tosc	—	7 Tosc	—	Timers in Sync mode

These parameters are characterized but not tested.

t Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

FIGURE 36-11: CAPTURE/COMPARE/PWM TIMINGS (CCP)

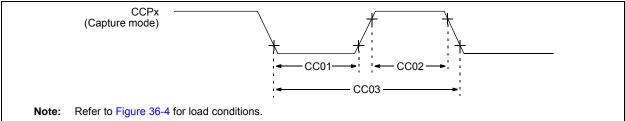


TABLE 36-13: CAPTURE/COMPARE/PWM REQUIREMENTS (CCP)

Standa	rd Oper	ating Conditions (unless	s otherwise stated	d)				
Param No.	Sym.	Character	stic	Min.	Тур†	Max.	Units	Conditions
CC01*	TccL	CCPx Input Low Time	No Prescaler	0.5Tcy + 20	_	_	ns	
			With Prescaler	20	_	_	ns	
CC02*	TccH	CCPx Input High Time	No Prescaler	0.5Tcy + 20	_	_	ns	
			With Prescaler	20			ns	
CC03*	TccP	CCPx Input Period		(3Tcy + 40)*N	_		ns	N = prescale value

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

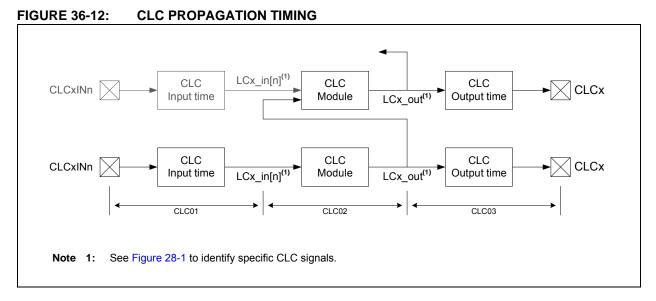


TABLE 36-14: CONFIGURATION LOGIC CELL (CLC) CHARACTERISTICS

	Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +125°C										
Param. No.	Sym.	Characteristic	Min.	Тур†	Max.	Units	Conditions				
CLC01*	TCLCIN	CLC input time	_	7	OS17	ns	(Note 1)				
CLC02*	TCLC	CLC module input to output progagation time		24 12	_	ns ns	VDD = 1.8V VDD > 3.6V				
CLC03*	TCLCOUT	CLC output time Rise Time		OS18		ns	(Note 1)				
		Fall Time	_	OS19	_	ns	(Note 1)				
CLC04*	FCLCMAX	CLC maximum switching frequency	_	45		MHz					

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: See Table 36-10 for OS17, OS18 and OS19 rise and fall times.

TABLE 36-15: ANALOG-TO-DIGITAL CONVERTER (ADC) CHARACTERISTICS^(1,2,3,4):

•	Operating Conditions (unless otherwise stated) VDD = 3.0V, TA = 25°C, Single-ended, 2 μs TAD, VREF+ = 3V, VREF- = Vss											
Param No.	Sym.	Characteristic	Min.	Тур†	Max.	Units	Conditions					
AD01	NR	Resolution		_	10	bit						
AD02	EIL	Integral Error	_	_	±1.7	LSb	VREF = 3.0V					
AD03	Edl	Differential Error	_	_	±1	LSb	No missing codes, VREF = 3.0V					
AD04	EOFF	Offset Error	_	_	±2.5	LSb	VREF = 3.0V					
AD05	Egn	Gain Error	_	_	±2.0	LSb	VREF = 3.0V					
AD06	VREF	Reference Voltage	1.8	_	Vdd	V	VREF = (VREF+ minus VREF-)					
AD07	VAIN	Full-Scale Range	Vss	_	VREF	V						
AD08	ZAIN	Recommended Impedance of Analog Voltage Source	—	—	10	kΩ	Can go higher if external 0.01µF capacitor is present on input pin.					

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Total Absolute Error includes integral, differential, offset and gain errors.

2: The ADC conversion result never decreases with an increase in the input voltage and has no missing codes.

3: ADC VREF is from external VREF+ pin, VDD pin or FVR, whichever is selected as reference input.

4: See Section 31.0 "DC and AC Characteristics Graphs and Charts" for operating characterization.

TABLE 36-16: ADC CONVERSION REQUIREMENTS

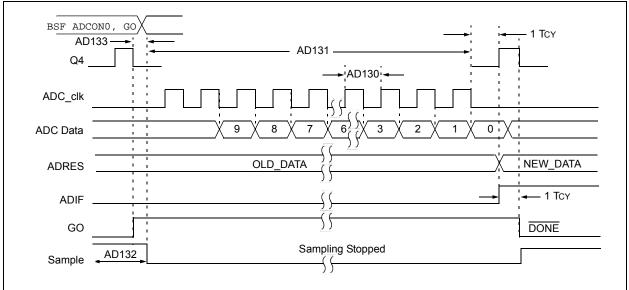
Standar	d Opei	rating Conditions (unless otherwise	e stated)				
Param No.	Sym.	Characteristic	Min.	Тур†	Max.	Units	Conditions
AD130*	TAD	ADC Clock Period (TADC)	1.0	—	9.0	μS	Fosc-based
		ADC Internal FRC Oscillator Period (TFRC)	1.0	2.5	6.0	μS	ADCS<1:0> = 11 (ADC FRC mode)
AD131	TCNV	Conversion Time (not including Acquisition Time) ⁽¹⁾	-	13	—	Tad	Set GO/DONE bit to conversion complete
AD132*	TACQ	Acquisition Time	_	5.0	_	μS	
AD133*	THCD	Holding Capacitor Disconnect Time	_	1/2 Tad			ADCS<2:0> \neq x11 (Fosc-based)
			_	1/2 TAD + 1TCY	_		ADCS<2:0> = $x11$ (FRC-based)

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The ADRES register may be read on the following TCY cycle.







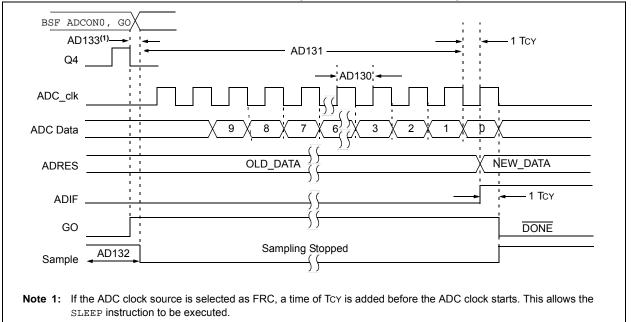


TABLE 36-17: OPERATIONAL AMPLIFIER (OPA)

Operating Conditions (unless otherwise stated) VDD = 3.0V, TA = 25°C, OPAxSP = 1 (High GBWP mode)										
Param No.	Symbol	Parameters	Min.	Тур.	Max.	Units	Conditions			
OPA01*	GBWP	Gain Bandwidth Product	—	3.5		MHz				
OPA02*	Ton	Turn-on Time	—	10		μS				
OPA03*	Рм	Phase Margin	—	40	_	degrees				
OPA04*	SR	Slew Rate	—	3		V/μs				
OPA05	Off	Offset	—	±3	±9	mV				
OPA06	CMRR	Common-Mode Rejection Ratio	52	70		dB				
OPA07*	Aol	Open Loop Gain	—	90		dB				
OPA08	VICM	Input Common-Mode Voltage	0	—	Vdd	V	VDD > 2.5V			
OPA09*	PSRR	Power Supply Rejection Ratio	—	80		dB				
OPA10*	HZ	High-Impedance On/Off Time	—	50	—	ns				

* These parameters are characterized but not tested.

TABLE 36-18: PROGRAMMABLE RAMP GENERATOR (PRG) SPECIFICATIONS

Operating Conditions (unless otherwise stated) VDD = 3.0V, TA = 25°C (unless otherwise stated)								
Param No.	Sym. Characteristics Min. Typ. Max. Units Comments							
PRG01	RRR	Rising Ramp Rate ⁽¹⁾	—	1	_	V/µs	PRGxCON2 = 10h	
PRG02	FRR	Falling Ramp Rate ⁽¹⁾	—	1	_	V/μs	PRGxCON2 = 10h	

* These parameters are characterized but not tested.

Note 1: Response time measured with one comparator input at VDD/2, while the other input transitions from Vss to VDD.

TABLE 36-19: COMPARATOR SPECIFICATIONS

Operating Conditions (unless otherwise stated) VDD = 3.0V, TA = 25°C See Section 31.0 "DC and AC Characteristics Graphs and Charts " for operating characterization.									
Sym.CharacteristicsMin.Typ.Max.UnitsComments									
VIOFF	Input Offset Voltage	—	±2.5	±5	mV	VICM = VDD/2			
VICM	Input Common-Mode Voltage	0		Vdd	V				
CMRR	Common-Mode Rejection Ratio	40	50	—	dB				
	Response Time Rising Edge	—	60	125	ns	CxSP = 1			
Trees(1)	Response Time Falling Edge		60	110	ns	CxSP = 1			
Tresp	Response Time Rising Edge		85		ns	CxSP = 0			
	Response Time Falling Edge		85		ns	CxSP = 0			
Тмс2о∨	Comparator Mode Change to Output Valid*	—	—	10	μS				
CHYSTER	Comparator Hysteresis	20	45	75	mV	CxHYS = 1			
	V, TA = 25° on 31.0 "D Sym. VIOFF VICM CMRR Tresp ⁽¹⁾ TMC2OV	V, TA = 25°C on 31.0 "DC and AC Characteristics Graph Sym. Characteristics VIOFF Input Offset Voltage VICM Input Common-Mode Voltage CMRR Common-Mode Rejection Ratio Response Time Rising Edge Response Time Falling Edge Response Time Rising Edge Response Time Rising Edge Tmc2ov Comparator Mode Change to Output Valid*	V, TA = 25°C On 31.0 "DC and AC Characteristics Graphs and C Sym. Characteristics Min. VIOFF Input Offset Voltage — VICM Input Common-Mode Voltage 0 CMRR Common-Mode Rejection Ratio 40 Response Time Rising Edge — Response Time Falling Edge — Response Time Rising Edge — Response Time Rising Edge — TMC2OV Comparator Mode Change to Output Valid*	V, TA = 25°Con 31.0 "DC and AC Characteristics Graphs and Charts" fSym.CharacteristicsMin.Typ.VIOFFInput Offset Voltage—±2.5VICMInput Common-Mode Voltage0—CMRRCommon-Mode Rejection Ratio4050Response Time Rising Edge—60Response Time Falling Edge—60Response Time Rising Edge—85TMC2OVComparator Mode Change to Output Valid*——	V, TA = 25°Con 31.0 "DC and AC Characteristics Graphs and Charts" for operationSym.CharacteristicsMin.Typ.Max.VIOFFInput Offset Voltage—±2.5±5VICMInput Common-Mode Voltage0—VDDCMRRCommon-Mode Rejection Ratio4050—CMRRCommon-Mode Rejection Ratio4050—Response Time Rising Edge—60125Response Time Falling Edge—60110Response Time Rising Edge—85—TMC2OVComparator Mode Change to Output Valid*——	V, TA = 25°Con 31.0 "DC and AC Characteristics Graphs and Charts" for operating characteristicsSym.CharacteristicsMin.Typ.Max.UnitsVIOFFInput Offset Voltage—±2.5±5mVVICMInput Common-Mode Voltage0—VDDVCMRRCommon-Mode Rejection Ratio4050—dBResponse Time Rising Edge—60125nsResponse Time Falling Edge—60110nsResponse Time Rising Edge—85—nsResponse Time Rising Edge—85—nsTMC2OVComparator Mode Change to Output Valid*——10µs			

* These parameters are characterized but not tested.

Note 1: Response time measured with one comparator input at VDD/2, while the other input transitions from Vss to VDD.

TABLE 36-20: 10-BIT DIGITAL-TO-ANALOG CONVERTER (DAC) SPECIFICATIONS

VDD = 3.0	V, TA = 2	ions (unless otherwise st 5°C " <mark>DC and AC Characteris</mark> t		hs and Cha	arts" for (operating	characterization.
Param No.	Sym.	Characteristics	Min.	Тур.	Max.	Units	Comments
DAC01*	CLSB	Step Size	_	VDD/1024	_	V	
DAC02	CINL	Integral Error ⁽²⁾		—	±2.0	LSb	For codes 0x004 to 0x3FB
DAC03	CDNL	Differential Error ⁽²⁾		_	± 1	LSb	
DAC04	COFF	Offset Error ⁽²⁾		_	± 3	LSb	
DAC05	Cgn	Gain Error ⁽²⁾		—	± 3	LSb	
DAC06*	CR	Unit Resistor Value (R)		300	_	Ω	
DAC07*	CST	Settling Time ⁽¹⁾		_	10	μS	

These parameters are characterized but not tested.

Note 1: Settling time measured while DACR<9:0> transitions from '0x000' to '0x3FF'.

2: Buffered by op amp in unity gain.

TABLE 36-21: 5-BIT DIGITAL-TO-ANALOG CONVERTER (DAC) SPECIFICATIONS

Operating Conditions (unless otherwise stated) VDD = 3.0V, TA = 25°C See Section 31.0 "DC and AC Characteristics Graphs and Charts" for operating characterization.										
Param No.	Sym. Characteristics Min. Ivp. Max. Units Comments									
DAC10*	CLSB	Step Size	_	VDD/32	_	V				
DAC11	CACC	Absolute Accuracy ⁽²⁾	_	—	± 0.5	LSb				
DAC12*	CR	Unit Resistor Value (R)	—	6000	—	Ω				
DAC13 [*] Cst Settling Time ⁽¹⁾ – – 10 μ s										

* These parameters are characterized but not tested.

Note 1: Settling time measured while DACR<4:0> transitions from '0x00' to '0x1F'.

2: Buffered by op amp in unity gain.

TABLE 36-22: ZERO CROSS PIN SPECIFICATIONS

Operating Conditions (unless otherwise stated) VDD = 3.0V, TA = 25°C									
Param. No.	Sym.	Characteristics	Min.	Тур.	Max.	Units	Comments		
ZC01	ZCPINV	Voltage on Zero Cross Pin		0.75	_	V			
ZC02	ZCDRV	Maximum source or sink current	_	_	600	μA			
ZC04	Zcisw	Response Time Rising Edge	_	1	—	μS			
		Response Time Falling Edge	_	1	—	μS			
ZC05	ZCOUT	Response Time Rising Edge		1	_	μS			
		Response Time Falling Edge	_	1	_	μS			

These parameters are characterized but not tested.

FIGURE 36-15: EUSART SYNCHRONOUS TRANSMISSION (MASTER/SLAVE) TIMING

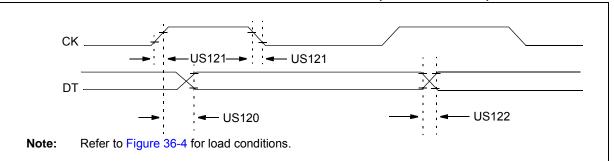


TABLE 36-23: EUSART SYNCHRONOUS TRANSMISSION REQUIREMENTS

Standard	d Operating C					
Param. No.	Symbol	Characteristic	Min.	Max.	Units	Conditions
US120	TCKH2DTV	SYNC XMIT (Master and Slave)		80	ns	$3.0V \le V\text{DD} \le 5.5V$
		Clock high to data-out valid	—	100	ns	$1.8V \leq V\text{DD} \leq 5.5V$
US121	TCKRF	Clock out rise time and fall time	_	45	ns	$3.0V \leq V\text{DD} \leq 5.5V$
		(Master mode)	_	50	ns	$1.8V \leq V\text{DD} \leq 5.5V$
US122	TDTRF	Data-out rise time and fall time	—	45	ns	$3.0V \leq V\text{DD} \leq 5.5V$
			_	50	ns	$1.8V \le V\text{DD} \le 5.5V$

FIGURE 36-16: EUSART SYNCHRONOUS RECEIVE (MASTER/SLAVE) TIMING

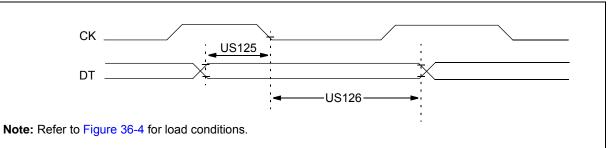


TABLE 36-24: EUSART SYNCHRONOUS RECEIVE REQUIREMENTS

Standar	Standard Operating Conditions (unless otherwise stated)							
Param. No.	Symbol	Characteristic	Min.	Max.	Units	Conditions		
US125	TDTV2CKL	<u>SYNC RCV (Master and Slave)</u> Data-setup before CK \downarrow (DT hold time)	10	_	ns			
US126	TCKL2DTL	Data-hold after CK \downarrow (DT hold time)	15	—	ns			

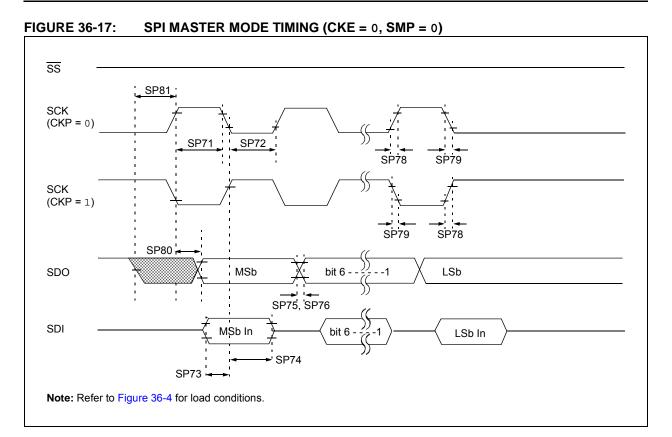
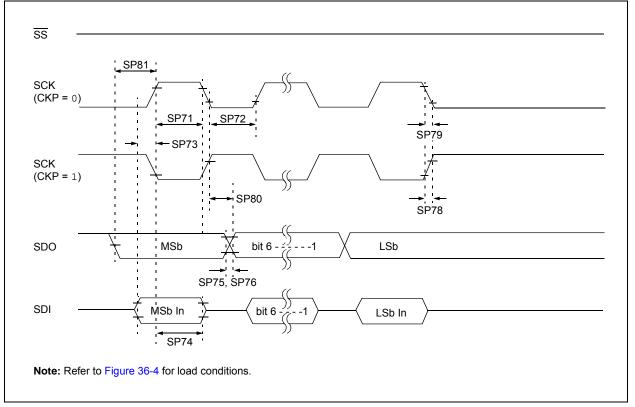


FIGURE 36-18: SPI MASTER MODE TIMING (CKE = 1, SMP = 1)



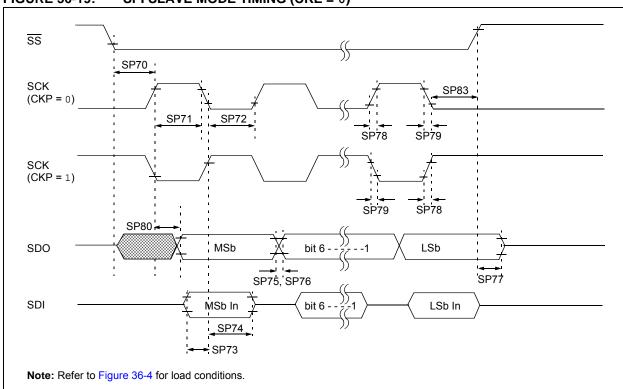


FIGURE 36-19: SPI SLAVE MODE TIMING (CKE = 0)



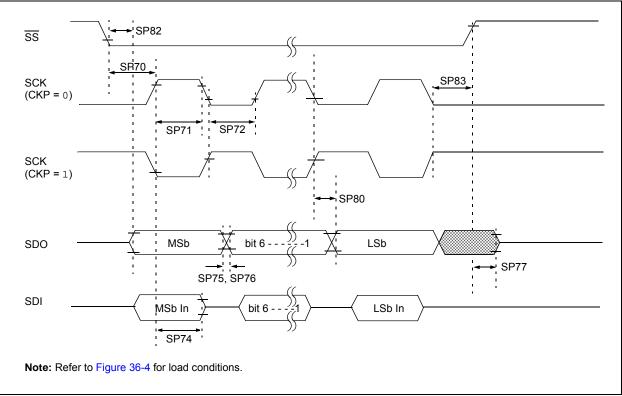


TABLE 36-25: SPI MODE REQUIREMENTS

Param No.	Symbol Characteristic		Min.	Тур†	Max.	Units	Conditions
SP70*	TssL2scH, TssL2scL	$\overline{\text{SS}}\downarrow$ to SCK \downarrow or SCK \uparrow input	2.25 TCY		_	ns	
SP71*	TscH	SCK input high time (Slave mode)	Tcy + 20	_	—	ns	
SP72*	TscL	SCK input low time (Slave mode)	Tcy + 20	_	—	ns	
SP73*	TDIV2scH, TDIV2scL	Setup time of SDI data input to SCK edge	100	_		ns	
SP74*	TscH2diL, TscL2diL	Hold time of SDI data input to SCK edge	100	_	—	ns	
SP75*	TDOR	SDO data output rise time	_	10	25	ns	$3.0V \le V\text{DD} \le 5.5^{\text{V}}$
			_	25	50	ns	$1.8V \le V\text{DD} \le 5.5^{\circ}$
SP76*	TDOF	SDO data output fall time	_	10	25	ns	
SP77*	TssH2doZ	SS↑ to SDO output high-impedance	10	_	50	ns	
SP78*	TscR	SCK output rise time	_	10	25	ns	$3.0V \le VDD \le 5.5V$
		(Master mode)	_	25	50	ns	$1.8V \le VDD \le 5.5V$
SP79*	TscF	SCK output fall time (Master mode)		10	25	ns	
SP80*	TscH2doV,	SDO data output valid after SCK	_	_	50	ns	$3.0V \le VDD \le 5.5V$
	TscL2DoV	edge	_	_	145	ns	$1.8V \le VDD \le 5.5V$
SP81*	TDOV2scH, TDOV2scL	SDO data output setup to SCK edge	1 Tcy	_		ns	
SP82*	TssL2doV	SDO data output valid after $\overline{\text{SS}}\downarrow$ edge	_	_	50	ns	
SP83*	TscH2ssH, TscL2ssH	SS ↑ after SCK edge	1.5 Tcy + 40	—	—	ns	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

FIGURE 36-21: I²C BUS START/STOP BITS TIMING

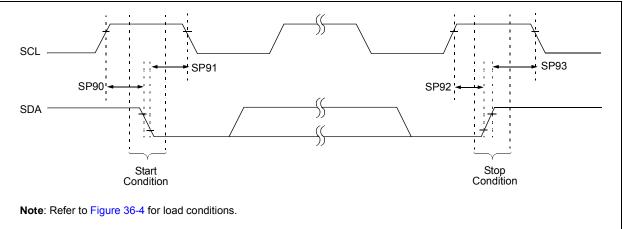
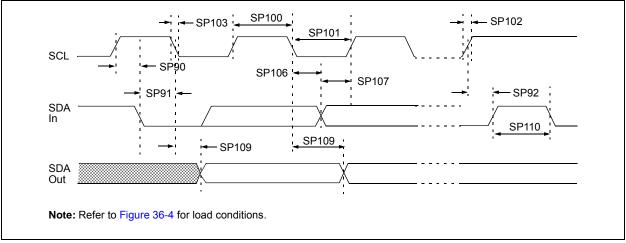


TABLE 36-26: I²C BUS START/STOP BITS REQUIREMENTS

Standard Operating Conditions (unless otherwise stated)									
Param No.	Symbol	Characteristic		Min.	Тур.	Max.	Units	Conditions	
SP90*	Tsu:sta	Start condition	100 kHz mode	4700	_	—	ns	Only relevant for Repeated	
		Setup time	400 kHz mode	600	_	—		Start condition	
SP91*	THD:STA	Start condition	100 kHz mode	4000	_	—	ns	After this period, the first	
		Hold time	400 kHz mode	600		—		clock pulse is generated	
SP92*	Tsu:sto	Stop condition	100 kHz mode	4700	_	—	ns		
		Setup time	400 kHz mode	600	—	—			
SP93	THD:STO	Stop condition	100 kHz mode	4000	—	—	ns		
		Hold time	400 kHz mode	600	_	_			

* These parameters are characterized but not tested.

FIGURE 36-22: I²C BUS DATA TIMING



Standard	Operating	Conditions (unless	otherwise stated	I)	÷	÷	
Param. No.	Symbol	Characte	eristic	Min.	Max.	Units	Conditions
SP100*	Тнідн	Clock high time	100 kHz mode	4.0	_	μS	Device must operate at a minimum of 1.5 MHz
			400 kHz mode	0.6	—	μS	Device must operate at a minimum of 10 MHz
			SSP module	1.5TCY	—		
SP101*	P101* TLOW Clock low time	100 kHz mode	4.7	—	μS	Device must operate at a minimum of 1.5 MHz	
			400 kHz mode	1.3	—	μS	Device must operate at a minimum of 10 MHz
			SSP module	1.5TCY	—		
SP102*	TR	SDA and SCL rise	100 kHz mode	_	1000	ns	
		time	400 kHz mode	20 + 0.1Св	300	ns	CB is specified to be from 10-400 pF
SP103*	TF	SDA and SCL fall	100 kHz mode	—	250	ns	
		time	400 kHz mode	20 + 0.1Св	250	ns	CB is specified to be from 10-400 pF
SP106*	THD:DAT	Data input hold time	100 kHz mode	0	—	ns	
			400 kHz mode	0	0.9	μs	
SP107*	TSU:DAT	Data input setup	100 kHz mode	250	—	ns	(Note 2)
		time	400 kHz mode	100	—	ns	
SP109*	ΤΑΑ	Output valid from	100 kHz mode	—	3500	ns	(Note 1)
		clock	400 kHz mode	_		ns	
SP110*	TBUF	Bus free time	100 kHz mode	4.7	_	μs	Time the bus must be free
			400 kHz mode	1.3	—	μS	before a new transmission can start
SP111	Св	Bus capacitive loadir	ng	_	400	pF	

* These parameters are characterized but not tested.

Note 1: As a transmitter, the device must provide this internal minimum delay time to bridge the undefined region (min. 300 ns) of the falling edge of SCL to avoid unintended generation of Start or Stop conditions.

2: A Fast mode (400 kHz) I²C bus device can be used in a Standard mode (100 kHz) I²C bus system, but the requirement Tsu:DAT \ge 250 ns must then be met. This will automatically be the case if the device does not stretch the low period of the SCL signal. If such a device does stretch the low period of the SCL signal, it must output the next data bit to the SDA line TR max. + TSU:DAT = 1000 + 250 = 1250 ns (according to the Standard mode I²C bus specification), before the SCL line is released.

NOTES:

37.0 DC AND AC CHARACTERISTICS GRAPHS AND CHARTS

The graphs and tables provided in this section are for **design guidance** and are **not tested**.

In some graphs or tables, the data presented are **outside specified operating range** (i.e., outside specified VDD range). This is for **information only** and devices are ensured to operate properly only within the specified range.

Unless otherwise noted, all graphs apply to both the L and LF devices.

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore, outside the warranted range.

"Typical" represents the mean of the distribution at 25°C. "Maximum", "Max.", "Minimum" or "Min." represents (mean + 3σ) or (mean - 3σ) respectively, where σ is a standard deviation, over each temperature range.

Note: Unless otherwise noted, VIN = 5V, Fosc = 300 kHz, CIN = 0.1 μ F, TA = 25°C.

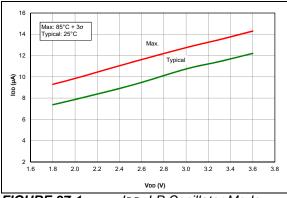


FIGURE 37-1: IDD, LP Oscillator Mode, Fosc = 32 kHz, PIC16LF1777/8/9 Only.

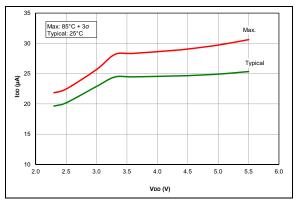


FIGURE 37-2: IDD, LP Oscillator Mode, Fosc = 32 kHz, PIC16F1777/8/9 Only.

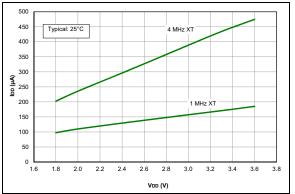


FIGURE 37-3: IDD Typical, XT and EXTRC Oscillator, PIC16LF1777/8/9 Only.

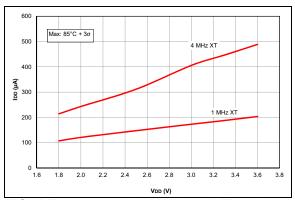


FIGURE 37-4: IDD Maximum, XT and EXTRC Oscillator, PIC16LF1777/8/9 Only.

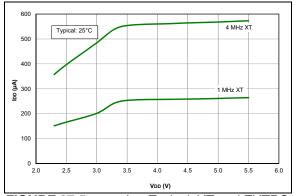
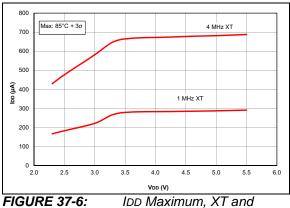


FIGURE 37-5: IDD Typical, XT and EXTRC Oscillator, PIC16F1777/8/9 Only.



EXTRC Oscillator, PIC16F1777/8/9 Only.

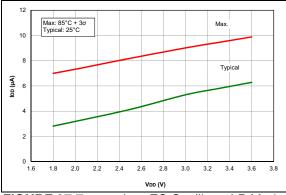


FIGURE 37-7: IDD, EC Oscillator LP Mode, Fosc = 32 kHz, PIC16LF1777/8/9 Only.

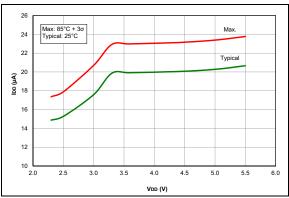


FIGURE 37-8: IDD, EC Oscillator LP Mode, Fosc = 32 kHz, PIC16F1777/8/9 Only.

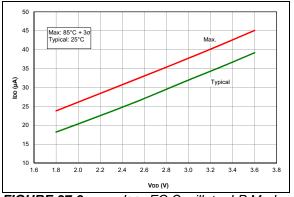


FIGURE 37-9: IDD, EC Oscillator LP Mode, Fosc = 500 kHz, PIC16LF1777/8/9 Only.

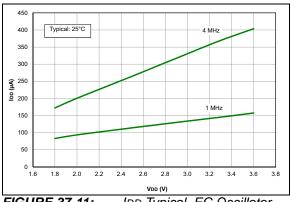


FIGURE 37-11: IDD Typical, EC Oscillator MP Mode, PIC16LF1777/8/9 Only.

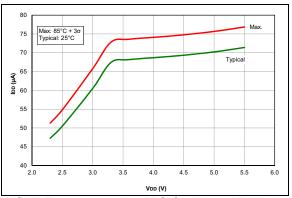


FIGURE 37-10: IDD, EC Oscillator LP Mode, Fosc = 500 kHz, PIC16F1777/8/9 Only.

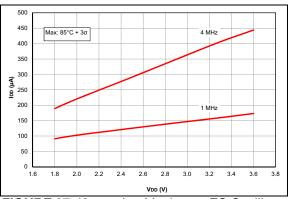


FIGURE 37-12: IDD Maximum, EC Oscillator MP Mode, PIC16LF1777/8/9 Only.

Note: Unless otherwise noted, VIN = 5V, Fosc = 300 kHz, CIN = 0.1 μ F, TA = 25°C.

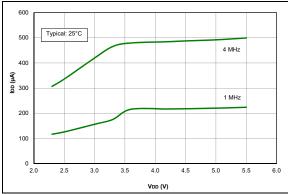


FIGURE 37-13: IDD Typical, EC Oscillator MP Mode, PIC16F1777/8/9 Only.

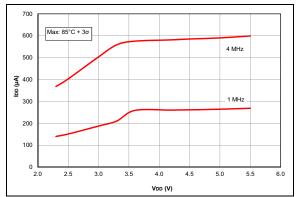


FIGURE 37-14: IDD Maximum, EC Oscillator MP Mode, PIC16F1777/8/9 Only.

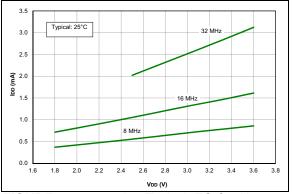


FIGURE 37-15: IDD Typical, EC Oscillator HP Mode, PIC16LF1777/8/9 Only.

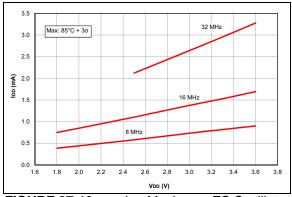


FIGURE 37-16: IDD Maximum, EC Oscillator HP Mode, PIC16LF1777/8/9 Only.

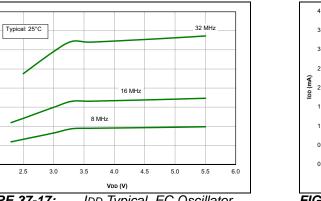


FIGURE 37-17: IDD Typical, EC Oscillator HP Mode, PIC16F1777/8/9 Only.

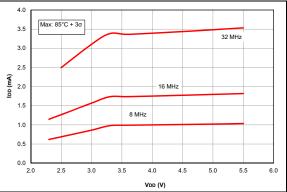


FIGURE 37-18: IDD Maximum, EC Oscillator HP Mode, PIC16F1777/8/9 Only.

4.0

3.5

3.0

2.5

1.5

1.0

0.5

0.0

2.0

(**F**m) 2.0

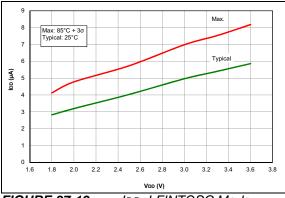


FIGURE 37-19: IDD, LFINTOSC Mode, Fosc = 31 kHz, PIC16LF1777/8/9 Only.

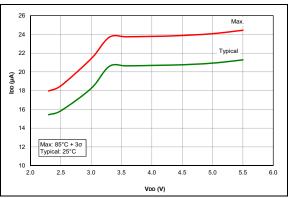


FIGURE 37-20: IDD, LFINTOSC Mode, Fosc = 31 kHz, PIC16F1777/8/9 Only.

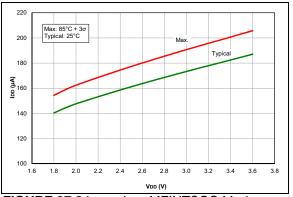
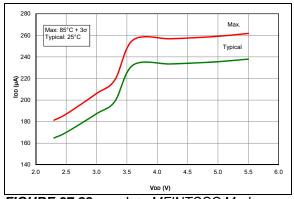
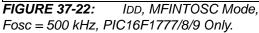


FIGURE 37-21: IDD, MFINTOSC Mode, Fosc = 500 kHz, PIC16LF1777/8/9 Only.





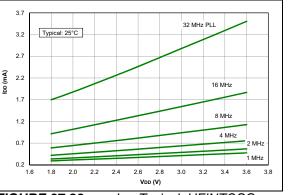


FIGURE 37-23: IDD Typical, HFINTOSC Mode, PIC16LF1777/8/9 Only.

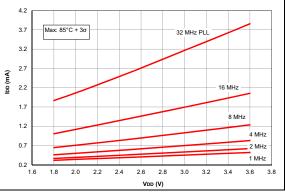


FIGURE 37-24: IDD Maximum, HFINTOSC Mode, PIC16LF1777/8/9 Only.

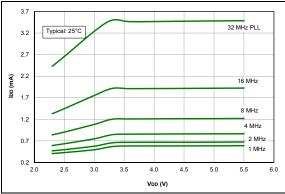


FIGURE 37-25: IDD Typical, HFINTOSC Mode, PIC16F1777/8/9 Only.

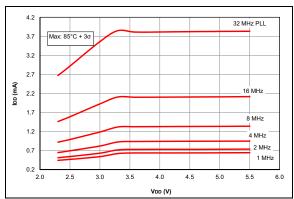


FIGURE 37-26: IDD Maximum, HFINTOSC Mode, PIC16F1777/8/9 Only.

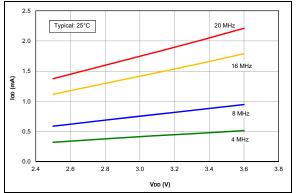


FIGURE 37-27: IDD Typical, HS Oscillator, 25°C, PIC16LF1777/8/9 Only.

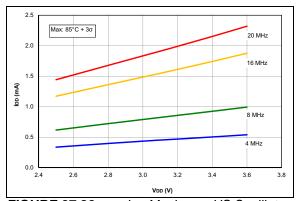


FIGURE 37-28: IDD Maximum, HS Oscillator, PIC16LF1777/8/9 Only.

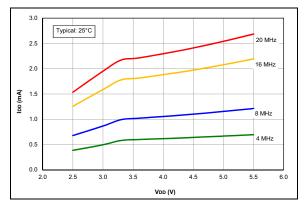


FIGURE 37-29: IDD Typical, HS Oscillator, 25°C, PIC16F1777/8/9 Only.

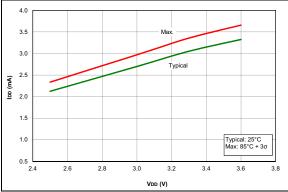


FIGURE 37-30: IDD, HS Oscillator (8 MHz + 4x PLL), PIC16LF1777/8/9 Only.

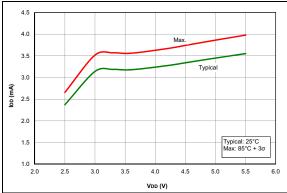


FIGURE 37-31: IDD, HS Oscillator, 32 MHz (8 MHz + 4x PLL), PIC16F1777/8/9 Only.

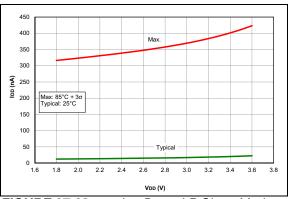


FIGURE 37-32: IPD Base, LP Sleep Mode, PIC16LF1777/8/9 Only.

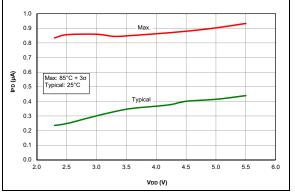
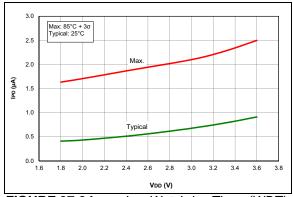
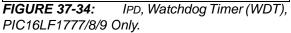


FIGURE 37-33: IPD Base, LP Sleep Mode (VREGPM = 1), PIC16F1777/8/9 Only.





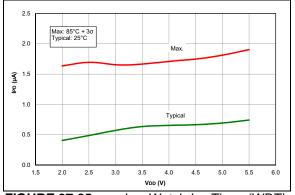
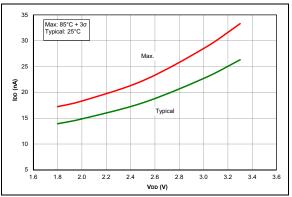
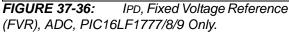


FIGURE 37-35: IPD, Watchdog Timer (WDT), PIC16F1777/8/9 Only.





Note: Unless otherwise noted, VIN = 5V, Fosc = 300 kHz, CIN = 0.1 µF, TA = 25°C.

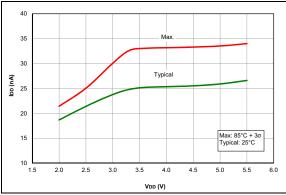


FIGURE 37-37: IPD, Fixed Voltage Reference (FVR), ADC, PIC16F1777/8/9 Only.

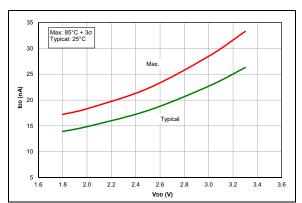


FIGURE 37-38: IPD, Fixed Voltage Reference (FVR), DAC/Comparator, PIC16LF1777/8/9 Only.

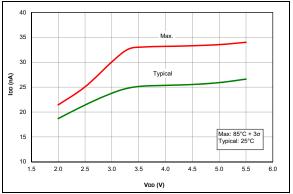


FIGURE 37-39: IPD, Fixed Voltage Reference (FVR), DAC/Comparator, PIC16F1777/8/9 Only.

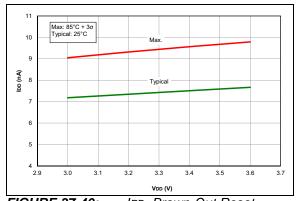
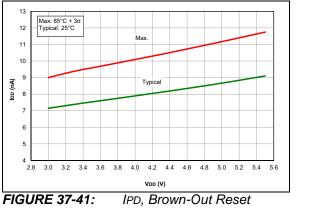
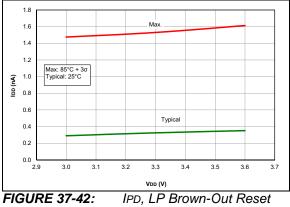


FIGURE 37-40: IPD, Brown-Out Reset (BOR), BORV = 1, PIC16LF1777/8/9 Only.



(BOR), BORV = 1, PIC16F1777/8/9 Only.



(LPBOR = 0), PIC16LF1777/8/9 Only.

(VU) aal

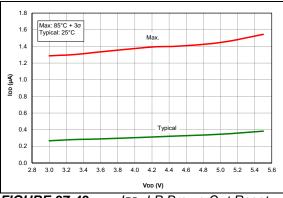


FIGURE 37-43: IPD, LP Brown-Out Reset (LPBOR = 0), PIC16F1777/8/9 Only.

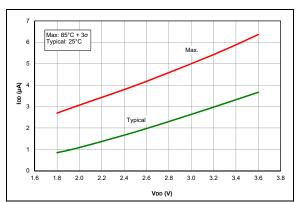


FIGURE 37-44: IPD, Timer1 Oscillator, Fosc = 32 kHz, PIC16LF1777/8/9 Only.

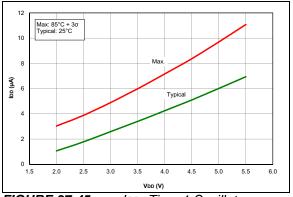
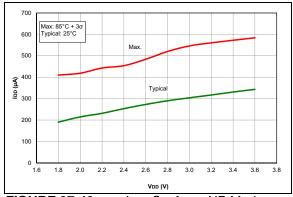
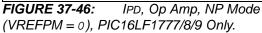
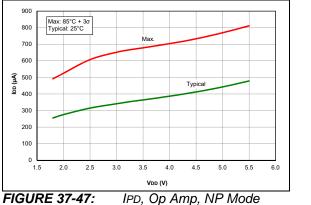


FIGURE 37-45: IPD, Timer1 Oscillator, Fosc = 32 kHz, PIC16F1777/8/9 Only.







(VREFPM = 0), PIC16F1777/8/9 Only.

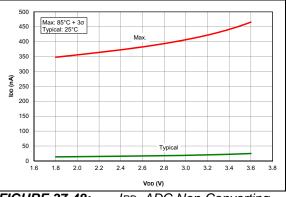


FIGURE 37-48: IPD, ADC Non-Converting, PIC16LF1777/8/9 Only.

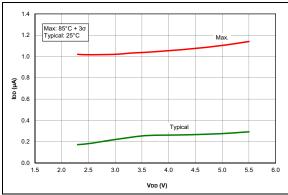


FIGURE 37-49: IPD, ADC Non-Converting, PIC16F1777/8/9 Only.

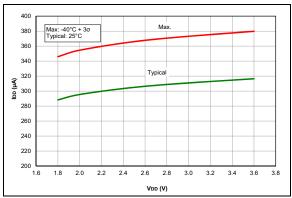


FIGURE 37-50: IPD, Comparator, NP Mode (VREGPM = 0), PIC16LF1777/8/9 Only.

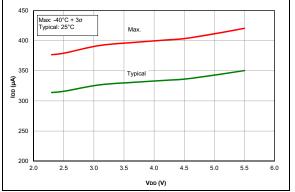
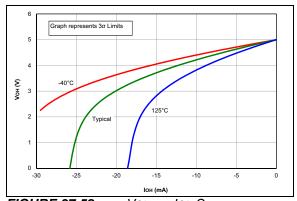
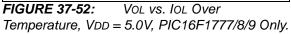


FIGURE 37-51: IPD, Comparator, NP Mode (VREGPM = 0), PIC16F1777/8/9 Only.





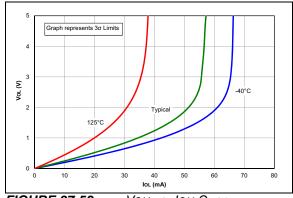


FIGURE 37-53: Voн vs. Ioн Over Temperature, VDD = 5.0V, PIC16F1777/8/9 Only.

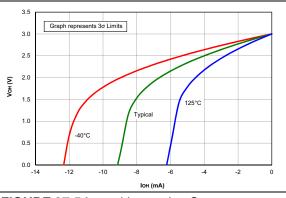


FIGURE 37-54: VOL vs. IOL Over Temperature, VDD = 3.0V.

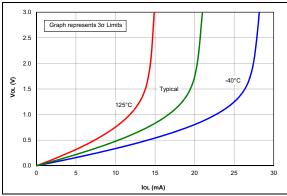


FIGURE 37-55: VOH vs. IOH Over Temperature, VDD = 3.0V.

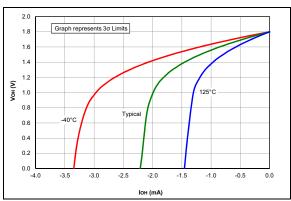


FIGURE 37-56: VOL vs. IOL Over Temperature, VDD = 1.8V, PIC16LF1777/8/9 Only.

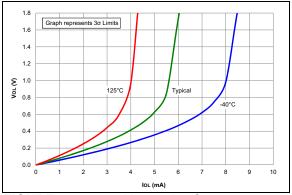


FIGURE 37-57: VOL vs. IOL Over Temperature, VDD = 1.8V, PIC16LF1777/8/9 Only.

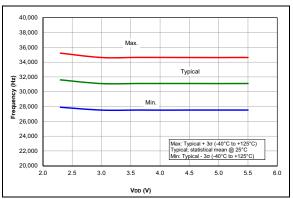


FIGURE 37-59: LFINTOSC Frequency, PIC16F1777/8/9 Only.

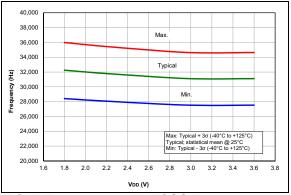


FIGURE 37-58: LFINTOSC Frequency, PIC16LF1777/8/9 Only.

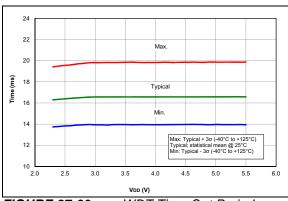


FIGURE 37-60: WDT Time-Out Period, PIC16F1777/8/9 Only.

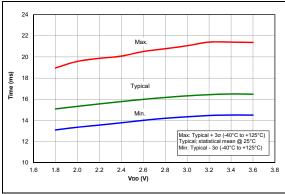


FIGURE 37-61: WDT Time-Out Period, PIC16LF1777/8/9 Only.

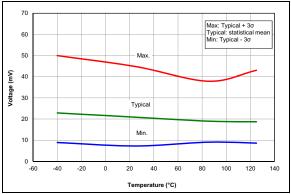


FIGURE 37-63: Brown-Out Reset Hysteresis, Low Trip Point (BORV = 1), PIC16LF1777/8/9 Only.

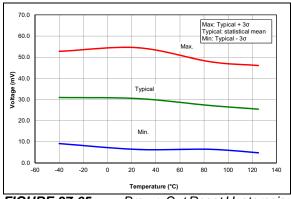


FIGURE 37-65: Brown-Out Reset Hysteresis, Low Trip Point (BORV = 1), PIC16F1773/6 Only.

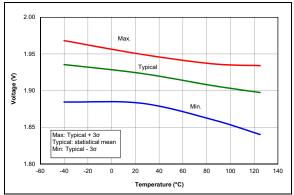


FIGURE 37-62: Brown-Out Reset Voltage, Low Trip Point (BORV = 1), PIC16LF1777/8/9 Only.

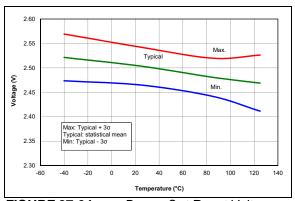


FIGURE 37-64: Brown-Out Reset Voltage, Low Trip Point (BORV = 1), PIC16F1773/6 Only.

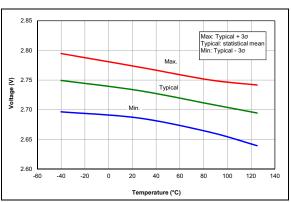


FIGURE 37-66: Brown-Out Reset Voltage, High Trip Point (BORV = 0).

Note: Unless otherwise noted, VIN = 5V, Fosc = 300 kHz, CIN = 0.1 µF, TA = 25°C.

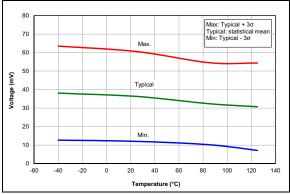
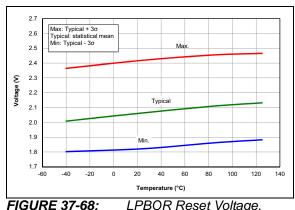
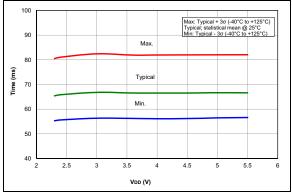


FIGURE 37-67: Brown-Out Reset Hysteresis, High Trip Point (BORV = 0).









PWRT Period, FIGURE 37-70: PIC16F1777/8/9 Only.

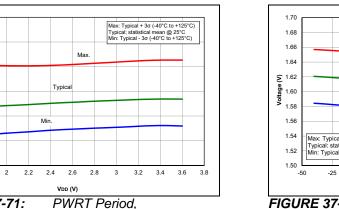


FIGURE 37-71: PIC16LF1773/6 Only.

100

90

80

60

50

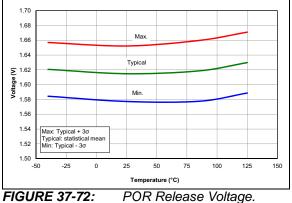
40

1.6

1.8

(ms)

Lime 70



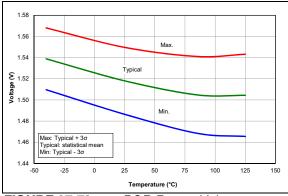


FIGURE 37-73: POR Rearm Voltage, NP Mode (VREGPM1 = 0), PIC16F1773/6 Only.

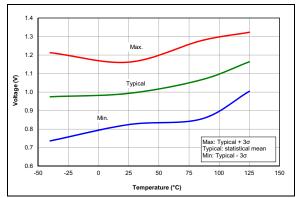


FIGURE 37-74: POR Rearm Voltage, NP Mode, PIC16LF1777/8/9 Only.

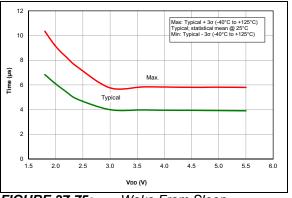


FIGURE 37-75: Wake From Sleep, VREGPM = 0.

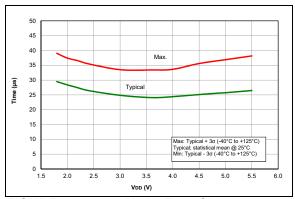


FIGURE 37-76: Wake From Sleep, VREGPM = 1.

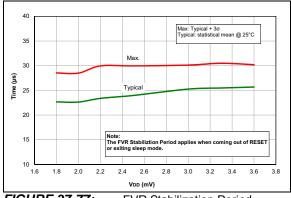
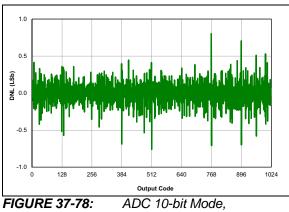
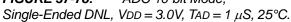


FIGURE 37-77: FVR Stabilization Period, PIC16LF1777/8/9 Only.





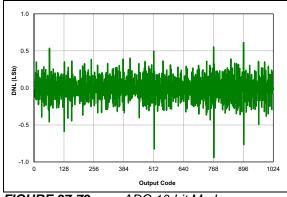


FIGURE 37-79: ADC 10-bit Mode, Single-Ended DNL, VDD = 3.0V, TAD = 4μ S, 25° C.

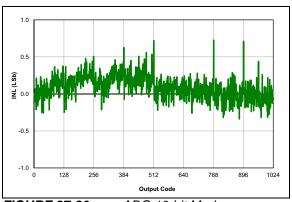


FIGURE 37-80: ADC 10-bit Mode, Single-Ended INL, VDD = 3.0V, TAD = 1μ S, 25° C.

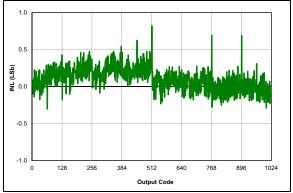


FIGURE 37-81: ADC 10-bit Mode, Single-Ended INL, VDD = 3.0V, TAD = 4μ S, 25° C.

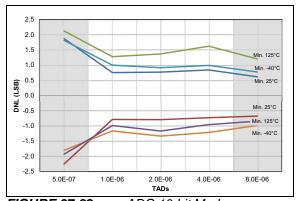


FIGURE 37-82: ADC 10-bit Mode, Single-Ended DNL, VDD = 3.0V, VREF = 3.0V.

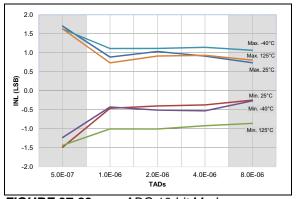
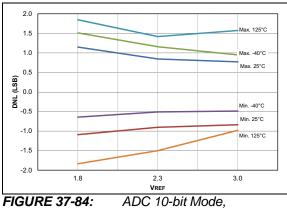


FIGURE 37-83: ADC 10-bit Mode, Single-Ended INL, VDD = 3.0V, VREF = 3.0V.



Single-Ended DNL, VDD = 3.0V, $TAD = 1 \ \mu S$.

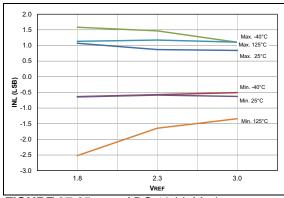


FIGURE 37-85: ADC 10-bit Mode, Single-Ended INL, VDD = 3.0V, TAD = 1μ S.

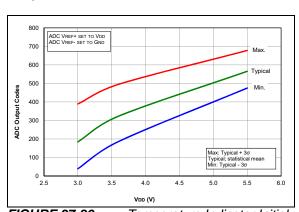


FIGURE 37-86: Temperature Indicator Initial Offset, High Range, Temp. = 20°C, PIC16F1777/8/9 only.

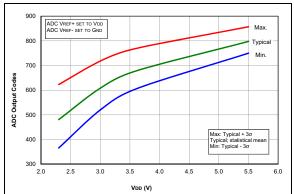


FIGURE 37-87: Temperature Indicator Initial Offset, Low Range, Temp. = 20°C, PIC16F1777/8/9 only.

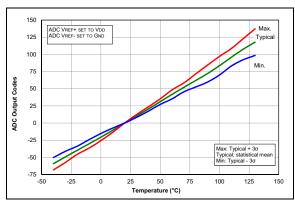


FIGURE 37-89: Temperature Indicator Slope Normalized to 20°C, PIC16F1777/8/9 only.

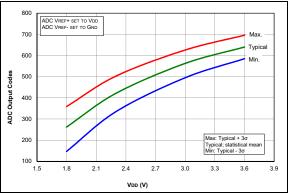


FIGURE 37-88: Temperature Indicator Initial Offset, Low Range, Temp. = 20°C, PIC16LF1773/6 only.

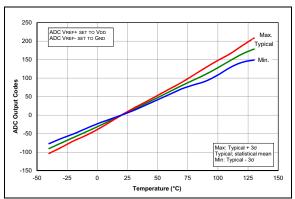


FIGURE 37-90: Temperature Indicator Slope Normalized to 20°C, High Range, VDD = 3.6V, PIC16F1777/8/9 only.

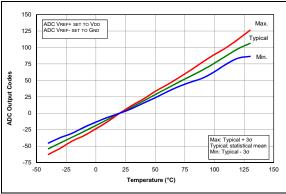


FIGURE 37-91: Temperature Indicator Slope Normalized to 20°C, Low Range, VDD = 3.0V, PIC16F1777/8/9 only.

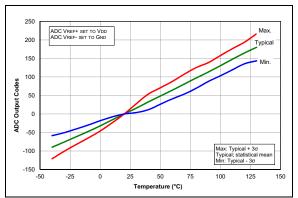


FIGURE 37-92: Temp. Indicator Slope Normalized to 20°C, Low Range, VDD = 1.8V, PIC16LF1773/6 Only.

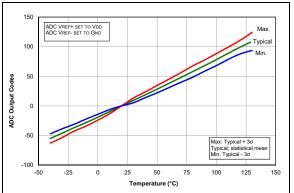


FIGURE 37-93: Temp. Indicator Slope Normalized to 20°C, Low Range, VDD = 3.0V, PIC16LF1773/6 Only.

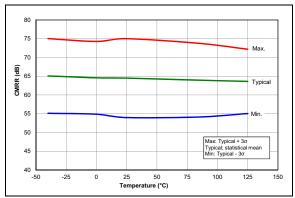


FIGURE 37-95: Op Amp, Common Mode Rejection Ratio (CMRR), VDD = 3.0V.

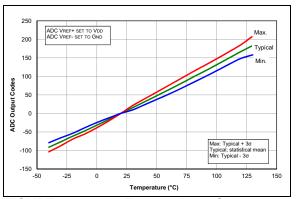
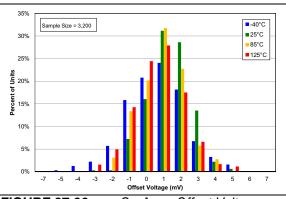
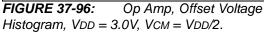


FIGURE 37-94: Temp. Indicator Slope Normalized to 20°C, High Range, VDD = 3.6V, PIC16LF1773/6 Only.





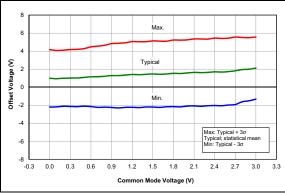


FIGURE 37-97: Op Amp, Offset over Common Mode Voltage, VDD = 3.0V, Temp. = 25°C

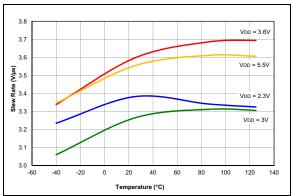


FIGURE 37-99: Op Amp, Output Slew Rate, Rising Edge, PIC16F1777/8/9 Only.

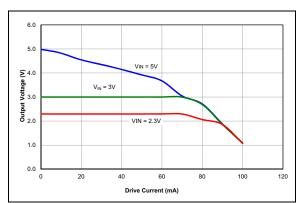


FIGURE 37-101: Op Amp, Output Drive Strength, VDD = 5.0V, Temp. = 25°C, PIC16F1777/8/9 Only.

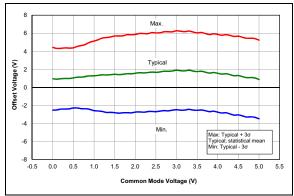


FIGURE 37-98: Op Amp, Offset over Common Mode Voltage, VDD = 5.0V, Temp. = 25°C, PIC16F1777/8/9 Only.

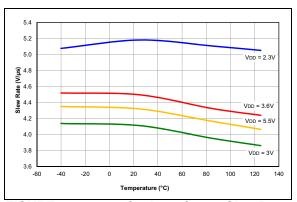


FIGURE 37-100: Op Amp, Output Slew Rate, Falling Edge, PIC16F1777/8/9 Only.

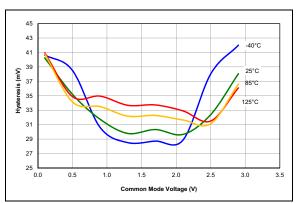


FIGURE 37-102: Comparator Hysteresis, NP Mode (CxSP = 1), VDD = 3.0V, Typical Measured Values.

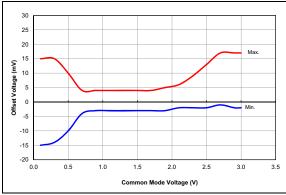


FIGURE 37-103: Comparator Offset, NP Mode (CxSP = 1), VDD = 3.0V, Typical Measured Values at 25°C.

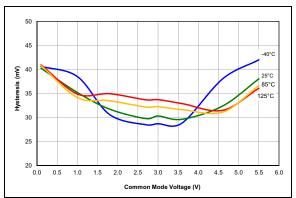


FIGURE 37-105: Comparator Hysteresis, NP Mode (CxSP = 1), VDD = 5.5V, Typical Measured Values, PIC16F1777/8/9 only.

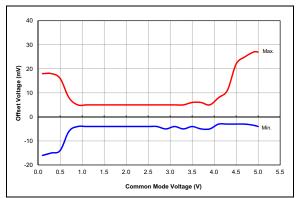


FIGURE 37-107: Comparator Offset, NP Mode (CxSP = 1), VDD = 5.5V, Typical Measured Values from -40°C to 125°C, PIC16F1777/8/9 only.

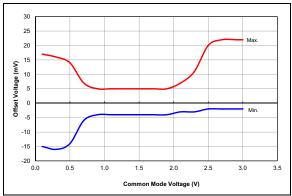


FIGURE 37-104: Comparator Offset, NP Mode (CxSP = 1), VDD = 3.0V, Typical Measured Values from -40°C to 125°C.

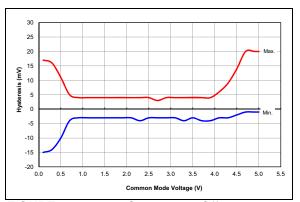


FIGURE 37-106: Comparator Offset, NP Mode (CxSP = 1), VDD = 5.0V, Typical Measured Values at 25°C, PIC16F1777/8/9 only.

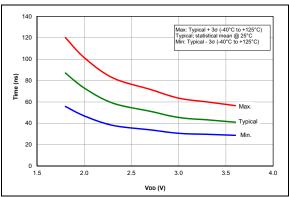


FIGURE 37-108: Comparator Response Time Over Voltage, NP Mode (CxSP = 1), Typical Measured Values.

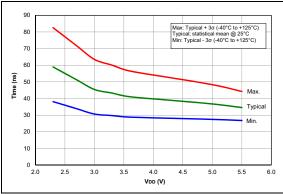


FIGURE 37-109: Comparator Response Time Over Voltage, NP Mode (CxSP = 1), Typical Measured Values, PIC16F1777/8/9 Only.

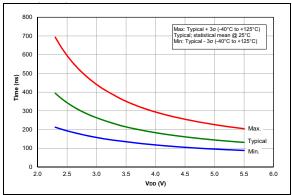


FIGURE 37-111: Comparator Output Filter Delay Time Over Temp., NP Mode (CxSP = 1), Typical Measured Values, PIC16F1777/8/9 Only.

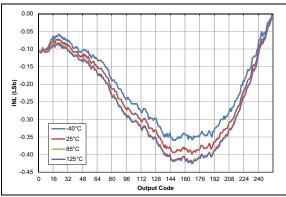


FIGURE 37-113: Typical DAC INL Error, VDD = 3.0V, VREF = External 3V.

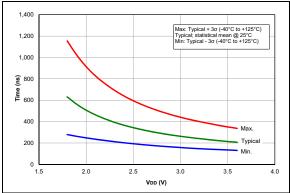


FIGURE 37-110: Comparator Output Filter Delay Time Over Temp., NP Mode (CxSP = 1), Typical Measured Values, PIC16LF1777/8/9 Only.

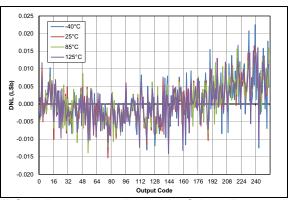


FIGURE 37-112: Typical DAC DNL Error, VDD = 3.0V, VREF = External 3V.

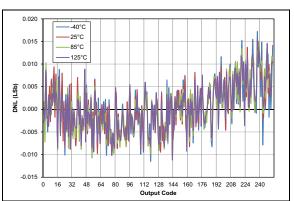


FIGURE 37-114: Typical DAC DNL Error, VDD = 5.0V, VREF = External 5V, PIC16F1777/8/9 Only.

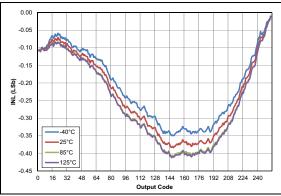


FIGURE 37-115: Typical DAC INL Error, VDD = 5.0V, VREF = External 5V, PIC16F1777/8/9 Only.

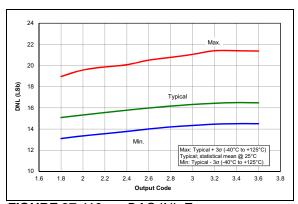


FIGURE 37-116: DAC INL Error, VDD = 3.0V, PIC16LF1773/6 Only.

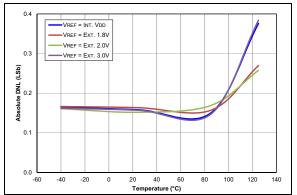


FIGURE 37-117: Absolute Value of DAC DNL Error, VDD = 3.0V, VREF = VDD.

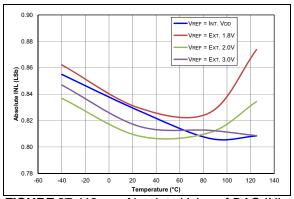


FIGURE 37-118: Absolute Value of DAC INL Error, VDD = 3.0V, VREF = VDD.

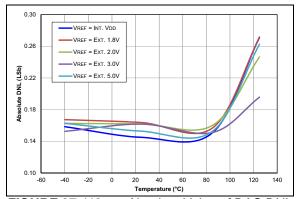


FIGURE 37-119: Absolute Value of DAC DNL Error, VDD = 5.0V, VREF = VDD, PIC16F1777/8/9 Only.

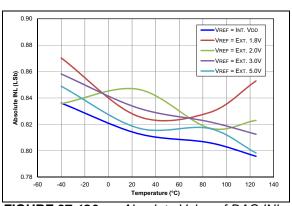


FIGURE 37-120: Absolute Value of DAC INL Error, VDD = 5.0V, VREF = VDD, PIC16F1777/8/9 Only.

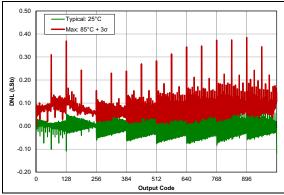


FIGURE 37-121: DAC DNL Error, VDD = 3.0V, VREF = External 3V.

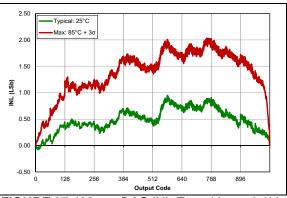


FIGURE 37-122: DAC INL Error, VDD = 3.0V, VREF = External 3V.

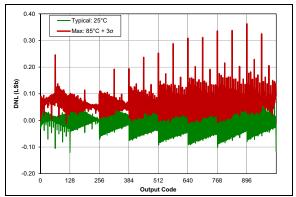


FIGURE 37-123: DAC DNL Error, VDD = 5.0V, VREF = External 5V, PIC16F1777/8/9 Only.

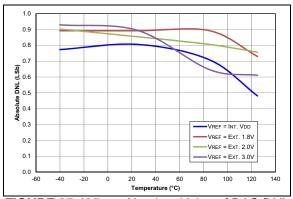


FIGURE 37-125: Absolute Value of DAC DNL Error, VDD = 3.0V, VREF = VDD.

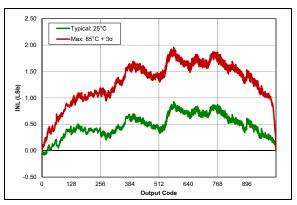


FIGURE 37-124: Typical DAC INL Error, VDD = 5.0V, VREF = External 5V, PIC16F1777/8/9 Only.

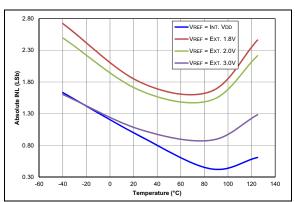


FIGURE 37-126: Absolute Value of DAC INL Error, VDD = 3.0V, VREF = VDD.

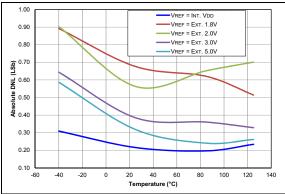


FIGURE 37-127: Absolute Value of DAC DNL Error, VDD = 5.0V, VREF = VDD, PIC16F1777/8/9 Only.

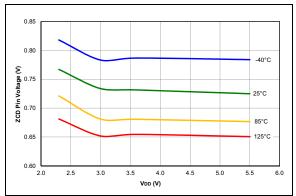


FIGURE 37-129: ZCD Pin Voltage, Typical Measured Values.

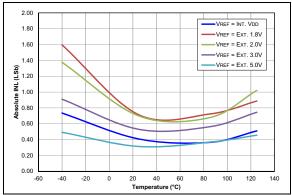


FIGURE 37-128: Absolute Value of DAC INL Error, VDD = 5.0V, VREF = VDD, PIC16F1777/8/9 Only.

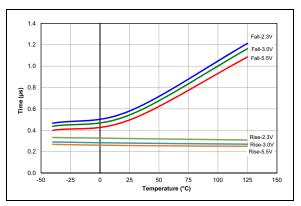


FIGURE 37-130: ZCD Response Time over Voltage, Typical Measured Values.

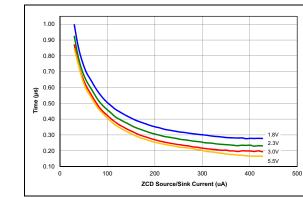


FIGURE 37-132: ZCD Pin Response Time over Current, Typical Measured Values from -40°C to 125°C.

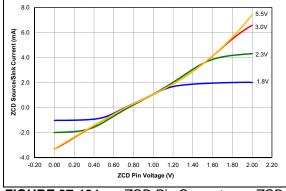


FIGURE 37-131: ZCD Pin Current over ZCD Pin Voltage, Typical Measured Values from -40°C to 125°C.

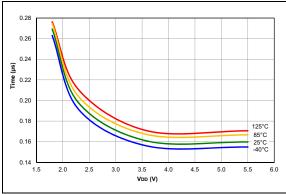


FIGURE 37-133: COG Dead-Band Delay, DBR/DBF = 32, Typical Measured Values.

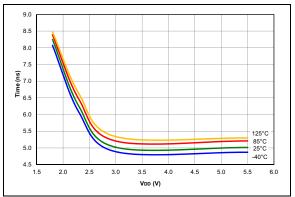


FIGURE 37-134: COG Dead-Band Delay, DBR/DBF Delay per Step, Typical Measured Values.

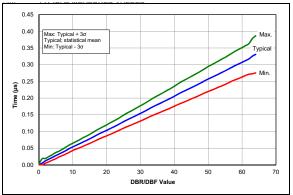


FIGURE 37-135: COG Dead-Band Delay per Step, Typical Measured Values.

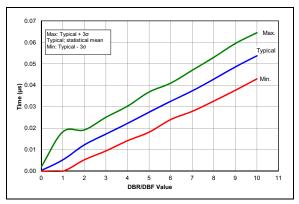


FIGURE 37-136: COG Dead-Band Delay per Step, Zoomed to First 10 Codes, Typical Measured Values.

38.0 DEVELOPMENT SUPPORT

The PIC[®] microcontrollers (MCU) and dsPIC[®] digital signal controllers (DSC) are supported with a full range of software and hardware development tools:

- Integrated Development Environment
- MPLAB[®] X IDE Software
- Compilers/Assemblers/Linkers
 - MPLAB XC Compiler
 - MPASM[™] Assembler
 - MPLINK[™] Object Linker/ MPLIB[™] Object Librarian
 - MPLAB Assembler/Linker/Librarian for Various Device Families
- Simulators
 - MPLAB X SIM Software Simulator
- · Emulators
 - MPLAB REAL ICE™ In-Circuit Emulator
- In-Circuit Debuggers/Programmers
 - MPLAB ICD 3
 - PICkit™ 3
- Device Programmers
 - MPLAB PM3 Device Programmer
- Low-Cost Demonstration/Development Boards, Evaluation Kits and Starter Kits
- Third-party development tools

38.1 MPLAB X Integrated Development Environment Software

The MPLAB X IDE is a single, unified graphical user interface for Microchip and third-party software, and hardware development tool that runs on Windows[®], Linux and Mac $OS^{®}$ X. Based on the NetBeans IDE, MPLAB X IDE is an entirely new IDE with a host of free software components and plug-ins for high-performance application development and debugging. Moving between tools and upgrading from software simulators to hardware debugging and programming tools is simple with the seamless user interface.

With complete project management, visual call graphs, a configurable watch window and a feature-rich editor that includes code completion and context menus, MPLAB X IDE is flexible and friendly enough for new users. With the ability to support multiple tools on multiple projects with simultaneous debugging, MPLAB X IDE is also suitable for the needs of experienced users.

Feature-Rich Editor:

- Color syntax highlighting
- Smart code completion makes suggestions and provides hints as you type
- Automatic code formatting based on user-defined rules
- · Live parsing

User-Friendly, Customizable Interface:

- Fully customizable interface: toolbars, toolbar buttons, windows, window placement, etc.
- · Call graph window
- Project-Based Workspaces:
- Multiple projects
- Multiple tools
- Multiple configurations
- · Simultaneous debugging sessions

File History and Bug Tracking:

- · Local file history feature
- Built-in support for Bugzilla issue tracker

38.2 MPLAB XC Compilers

The MPLAB XC Compilers are complete ANSI C compilers for all of Microchip's 8, 16, and 32-bit MCU and DSC devices. These compilers provide powerful integration capabilities, superior code optimization and ease of use. MPLAB XC Compilers run on Windows, Linux or MAC OS X.

For easy source level debugging, the compilers provide debug information that is optimized to the MPLAB X IDE.

The free MPLAB XC Compiler editions support all devices and commands, with no time or memory restrictions, and offer sufficient code optimization for most applications.

MPLAB XC Compilers include an assembler, linker and utilities. The assembler generates relocatable object files that can then be archived or linked with other relocatable object files and archives to create an executable file. MPLAB XC Compiler uses the assembler to produce its object file. Notable features of the assembler include:

- · Support for the entire device instruction set
- Support for fixed-point and floating-point data
- Command-line interface
- · Rich directive set
- Flexible macro language
- MPLAB X IDE compatibility

38.3 MPASM Assembler

The MPASM Assembler is a full-featured, universal macro assembler for PIC10/12/16/18 MCUs.

The MPASM Assembler generates relocatable object files for the MPLINK Object Linker, Intel[®] standard HEX files, MAP files to detail memory usage and symbol reference, absolute LST files that contain source lines and generated machine code, and COFF files for debugging.

The MPASM Assembler features include:

- Integration into MPLAB X IDE projects
- User-defined macros to streamline assembly code
- Conditional assembly for multipurpose source files
- Directives that allow complete control over the assembly process

38.4 MPLINK Object Linker/ MPLIB Object Librarian

The MPLINK Object Linker combines relocatable objects created by the MPASM Assembler. It can link relocatable objects from precompiled libraries, using directives from a linker script.

The MPLIB Object Librarian manages the creation and modification of library files of precompiled code. When a routine from a library is called from a source file, only the modules that contain that routine will be linked in with the application. This allows large libraries to be used efficiently in many different applications.

The object linker/library features include:

- Efficient linking of single libraries instead of many smaller files
- Enhanced code maintainability by grouping related modules together
- Flexible creation of libraries with easy module listing, replacement, deletion and extraction

38.5 MPLAB Assembler, Linker and Librarian for Various Device Families

MPLAB Assembler produces relocatable machine code from symbolic assembly language for PIC24, PIC32 and dsPIC DSC devices. MPLAB XC Compiler uses the assembler to produce its object file. The assembler generates relocatable object files that can then be archived or linked with other relocatable object files and archives to create an executable file. Notable features of the assembler include:

- · Support for the entire device instruction set
- · Support for fixed-point and floating-point data
- Command-line interface
- · Rich directive set
- Flexible macro language
- MPLAB X IDE compatibility

38.6 MPLAB X SIM Software Simulator

The MPLAB X SIM Software Simulator allows code development in a PC-hosted environment by simulating the PIC MCUs and dsPIC DSCs on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a comprehensive stimulus controller. Registers can be logged to files for further run-time analysis. The trace buffer and logic analyzer display extend the power of the simulator to record and track program execution, actions on I/O, most peripherals and internal registers.

The MPLAB X SIM Software Simulator fully supports symbolic debugging using the MPLAB XC Compilers, and the MPASM and MPLAB Assemblers. The software simulator offers the flexibility to develop and debug code outside of the hardware laboratory environment, making it an excellent, economical software development tool.

38.7 MPLAB REAL ICE In-Circuit Emulator System

The MPLAB REAL ICE In-Circuit Emulator System is Microchip's next generation high-speed emulator for Microchip Flash DSC and MCU devices. It debugs and programs all 8, 16 and 32-bit MCU, and DSC devices with the easy-to-use, powerful graphical user interface of the MPLAB X IDE.

The emulator is connected to the design engineer's PC using a high-speed USB 2.0 interface and is connected to the target with either a connector compatible with in-circuit debugger systems (RJ-11) or with the new high-speed, noise tolerant, Low-Voltage Differential Signal (LVDS) interconnection (CAT5).

The emulator is field upgradeable through future firmware downloads in MPLAB X IDE. MPLAB REAL ICE offers significant advantages over competitive emulators including full-speed emulation, run-time variable watches, trace analysis, complex breakpoints, logic probes, a ruggedized probe interface and long (up to three meters) interconnection cables.

38.8 MPLAB ICD 3 In-Circuit Debugger System

The MPLAB ICD 3 In-Circuit Debugger System is Microchip's most cost-effective, high-speed hardware debugger/programmer for Microchip Flash DSC and MCU devices. It debugs and programs PIC Flash microcontrollers and dsPIC DSCs with the powerful, yet easy-to-use graphical user interface of the MPLAB IDE.

The MPLAB ICD 3 In-Circuit Debugger probe is connected to the design engineer's PC using a highspeed USB 2.0 interface and is connected to the target with a connector compatible with the MPLAB ICD 2 or MPLAB REAL ICE systems (RJ-11). MPLAB ICD 3 supports all MPLAB ICD 2 headers.

38.9 PICkit 3 In-Circuit Debugger/ Programmer

The MPLAB PICkit 3 allows debugging and programming of PIC and dsPIC Flash microcontrollers at a most affordable price point using the powerful graphical user interface of the MPLAB IDE. The MPLAB PICkit 3 is connected to the design engineer's PC using a fullspeed USB interface and can be connected to the target via a Microchip debug (RJ-11) connector (compatible with MPLAB ICD 3 and MPLAB REAL ICE). The connector uses two device I/O pins and the Reset line to implement in-circuit debugging and In-Circuit Serial Programming[™] (ICSP[™]).

38.10 MPLAB PM3 Device Programmer

The MPLAB PM3 Device Programmer is a universal, CE compliant device programmer with programmable voltage verification at VDDMIN and VDDMAX for maximum reliability. It features a large LCD display (128 x 64) for menus and error messages, and a modular, detachable socket assembly to support various package types. The ICSP cable assembly is included as a standard item. In Stand-Alone mode, the MPLAB PM3 Device Programmer can read, verify and program PIC devices without a PC connection. It can also set code protection in this mode. The MPLAB PM3 connects to the host PC via an RS-232 or USB cable. The MPLAB PM3 has high-speed communications and optimized algorithms for quick programming of large memory devices, and incorporates an MMC card for file storage and data applications.

38.11 Demonstration/Development Boards, Evaluation Kits, and Starter Kits

A wide variety of demonstration, development and evaluation boards for various PIC MCUs and dsPIC DSCs allows quick application development on fully functional systems. Most boards include prototyping areas for adding custom circuitry and provide application firmware and source code for examination and modification.

The boards support a variety of features, including LEDs, temperature sensors, switches, speakers, RS-232 interfaces, LCD displays, potentiometers and additional EEPROM memory.

The demonstration and development boards can be used in teaching environments, for prototyping custom circuits and for learning about various microcontroller applications.

In addition to the PICDEM[™] and dsPICDEM[™] demonstration/development board series of circuits, Microchip has a line of evaluation kits and demonstration software for analog filter design, KEELOQ[®] security ICs, CAN, IrDA[®], PowerSmart battery management, SEEVAL[®] evaluation system, Sigma-Delta ADC, flow rate sensing, plus many more.

Also available are starter kits that contain everything needed to experience the specified device. This usually includes a single application and debug capability, all on one board.

Check the Microchip web page (www.microchip.com) for the complete list of demonstration, development and evaluation kits.

38.12 Third-Party Development Tools

Microchip also offers a great collection of tools from third-party vendors. These tools are carefully selected to offer good value and unique functionality.

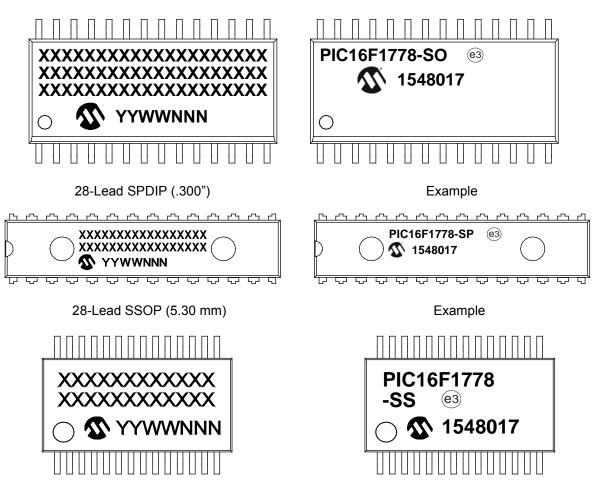
- Device Programmers and Gang Programmers from companies, such as SoftLog and CCS
- Software Tools from companies, such as Gimpel and Trace Systems
- Protocol Analyzers from companies, such as Saleae and Total Phase
- Demonstration Boards from companies, such as MikroElektronika, Digilent[®] and Olimex
- Embedded Ethernet Solutions from companies, such as EZ Web Lynx, WIZnet and IPLogika[®]

Example

39.0 PACKAGING INFORMATION

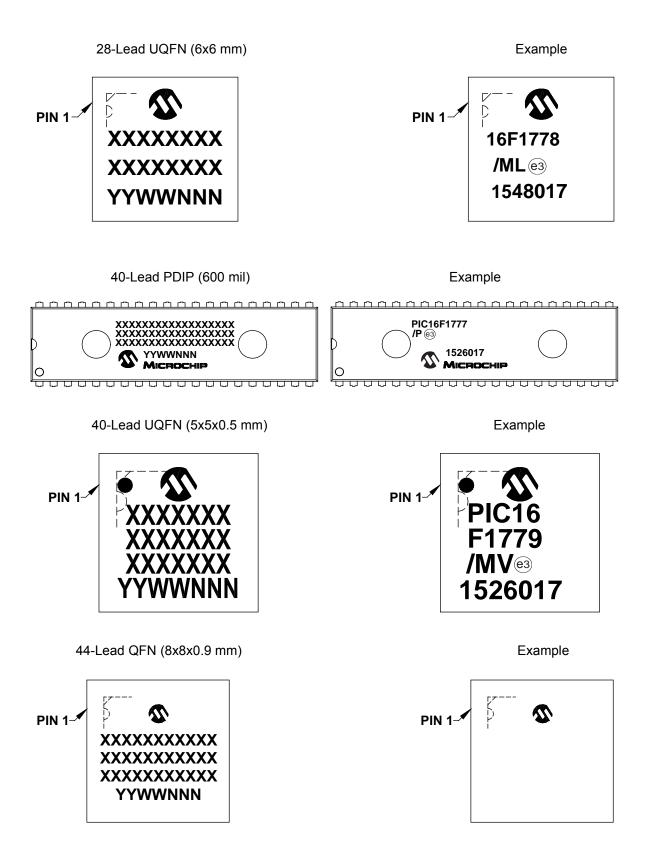
39.1 Package Marking Information

28-Lead SOIC (7.50 mm)



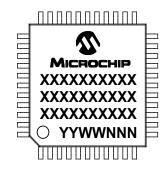
Legend	: XXX Y YY WW NNN @3 *	Customer-specific information Year code (last digit of calendar year) Year code (last 2 digits of calendar year) Week code (week of January 1 is week '01') Alphanumeric traceability code Pb-free JEDEC [®] designator for Matte Tin (Sn) This package is Pb-free. The Pb-free JEDEC [®] designator _(e3)) can be found on the outer packaging for this package.		
	In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.			

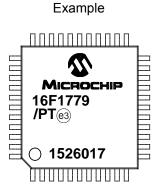
Package Marking Information (Continued)



Package Marking Information (Continued)

44-Lead TQFP (10x10x1 mm)



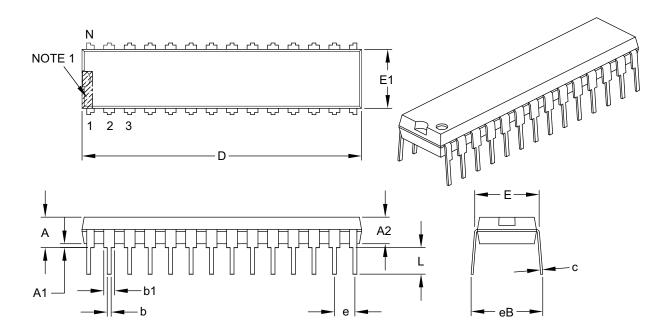


39.2 Package Details

The following sections give the technical details of the packages.

28-Lead Skinny Plastic Dual In-Line (SP) – 300 mil Body [SPDIP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		INCHES		
	Dimension Limits	MIN	NOM	MAX
Number of Pins	N 28			
Pitch	е	.100 BSC		
Top to Seating Plane	A	-	-	.200
Molded Package Thickness	A2	.120	.135	.150
Base to Seating Plane	A1	.015	-	-
Shoulder to Shoulder Width	E	.290	.310	.335
Molded Package Width	E1	.240	.285	.295
Overall Length	D	1.345	1.365	1.400
Tip to Seating Plane	L	.110	.130	.150
Lead Thickness	С	.008	.010	.015
Upper Lead Width	b1	.040	.050	.070
Lower Lead Width	b	.014	.018	.022
Overall Row Spacing §	eB	_	-	.430

Notes:

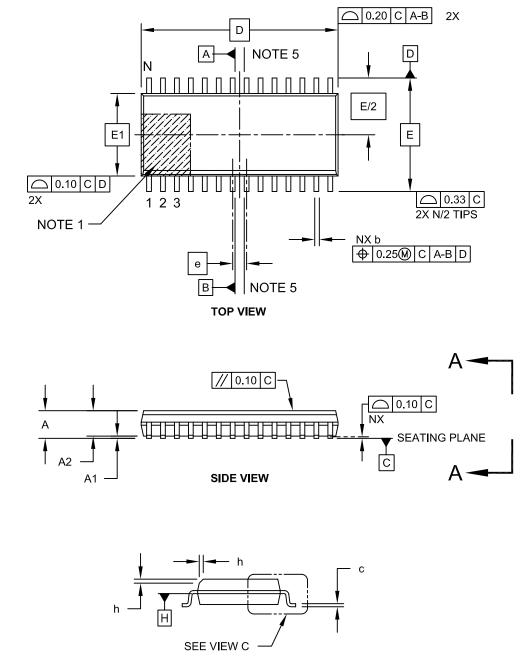
- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. § Significant Characteristic.
- 3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" per side.
- 4. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-070B

28-Lead Plastic Small Outline (SO) - Wide, 7.50 mm Body [SOIC]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging

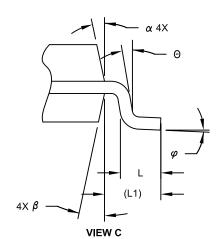


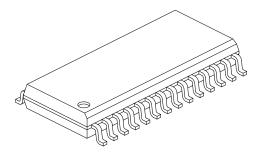
VIEW A-A

Microchip Technology Drawing C04-052C Sheet 1 of 2

28-Lead Plastic Small Outline (SO) - Wide, 7.50 mm Body [SOIC]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging





Units		MILLIMETERS		
Limits	MIN	NOM	MAX	
N		28		
е		1.27 BSC		
A	-	-	2.65	
A2	2.05	-	-	
A1	0.10	-	0.30	
E	10.30 BSC			
E1	7.50 BSC			
D	17.90 BSC			
h	0.25	-	0.75	
L	0.40	-	1.27	
L1	1.40 REF			
Θ	0°	-	-	
φ	0°	-	8°	
С	0.18	-	0.33	
b	0.31	-	0.51	
α	5°	-	15°	
β	5°	-	15°	
	Limits N e A A 2 A1 E E 1 D h L L 1 Ο 0 Ψ C C b b α	Limits MIN N $ A$ $ A2$ 2.05 $A1$ 0.10 E $ E1$ $ D$ $ D$ $ D$ $ C$ 0.40 $L1$ $ \Theta$ 0° φ 0° C 0.18 b 0.311 α 5°	Limits MIN NOM N 28 e 1.27 BSC A - A2 2.05 A1 0.10 E 10.30 BSC E1 7.50 BSC D 17.90 BSC h 0.25 L 0.40 L1 1.40 REF Θ 0° φ 0° c 0.18 b 0.31 α 5°	

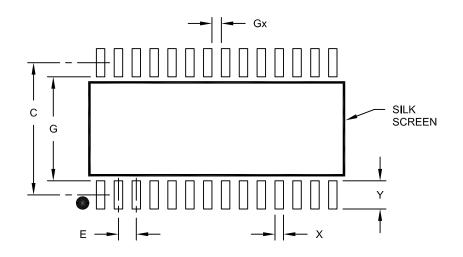
Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. § Significant Characteristic
- 3. Dimension D does not include mold flash, protrusions or gate burrs, which shall not exceed 0.15 mm per end. Dimension E1 does not include interlead flash or protrusion, which shall not exceed 0.25 mm per side.
- Dimensioning and tolerancing per ASME Y14.5M BSC: Basic Dimension. Theoretically exact value shown without tolerances. REF: Reference Dimension, usually without tolerance, for information purposes only.
- 5. Datums A & B to be determined at Datum H.

Microchip Technology Drawing C04-052C Sheet 2 of 2

28-Lead Plastic Small Outline (SO) - Wide, 7.50 mm Body [SOIC]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E	1.27 BSC		
Contact Pad Spacing	С		9.40	
Contact Pad Width (X28)	Х			0.60
Contact Pad Length (X28)	Y			2.00
Distance Between Pads	Gx	0.67		
Distance Between Pads	G	7.40		

Notes:

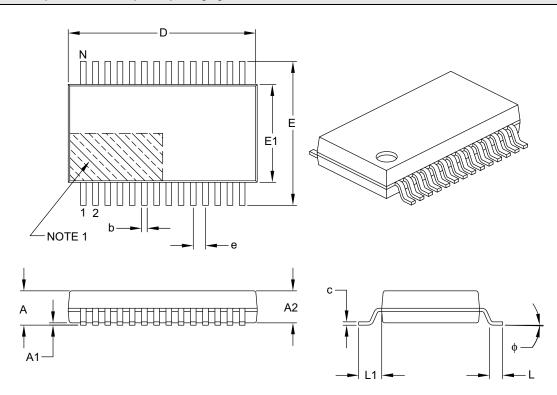
1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2052A

28-Lead Plastic Shrink Small Outline (SS) – 5.30 mm Body [SSOP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Number of Pins N		28		
Pitch	е		0.65 BSC	
Overall Height	А	_	-	2.00
Molded Package Thickness	A2	1.65	1.75	1.85
Standoff	A1	0.05	-	-
Overall Width	Е	7.40	7.80	8.20
Molded Package Width	E1	5.00	5.30	5.60
Overall Length	D	9.90	10.20	10.50
Foot Length	L	0.55	0.75	0.95
Footprint	L1 1.25 REF			
Lead Thickness	с	0.09	-	0.25
Foot Angle	¢	0°	4°	8°
Lead Width	b	0.22	-	0.38

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

2. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.20 mm per side.

3. Dimensioning and tolerancing per ASME Y14.5M.

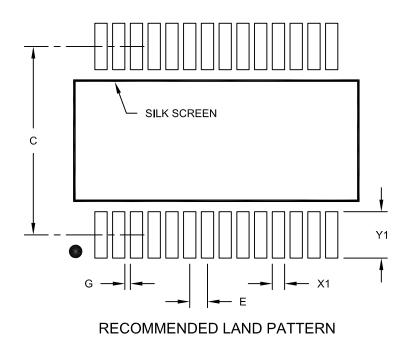
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-073B

28-Lead Plastic Shrink Small Outline (SS) - 5.30 mm Body [SSOP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	MILLIMETERS			
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E	0.65 BSC		
Contact Pad Spacing	С		7.20	
Contact Pad Width (X28)	X1			0.45
Contact Pad Length (X28)	Y1			1.75
Distance Between Pads	G	0.20		

Notes:

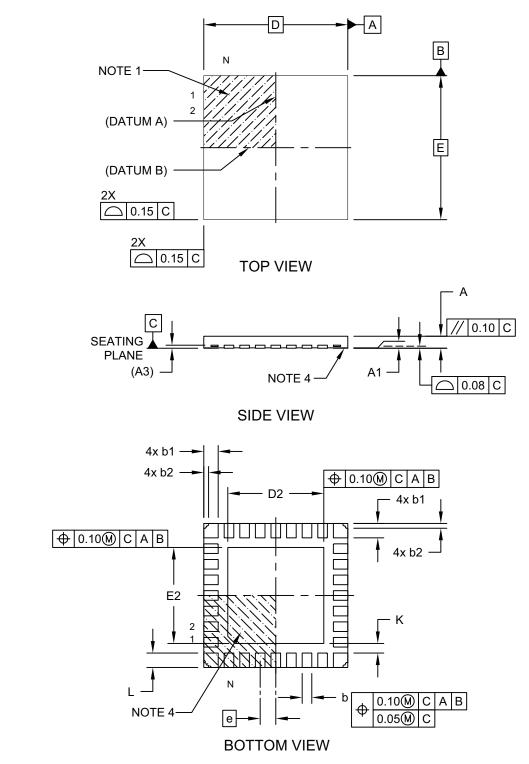
1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2073A

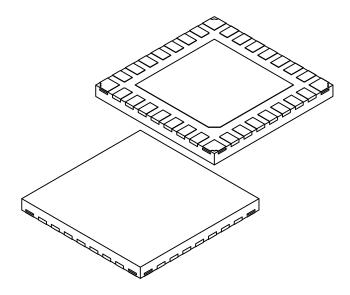
28-Lead Plastic Quad Flat, No Lead Package (MX) - 6x6x0.5mm Body [UQFN] Ultra-Thin with 0.40 x 0.60 mm Terminal Width/Length and Corner Anchors

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Microchip Technology Drawing C04-0209 Rev C Sheet 1 of 2

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units		MILLIMETERS		
Dimension	Limits	MIN	NOM	MAX	
Number of Pins	N		28		
Pitch	е		0.65 BSC		
Overall Height	Α	0.40	0.50	0.60	
Standoff	A1	0.00	0.02	0.05	
Terminal Thickness	(A3)	0.127 REF			
Overall Width	E	6.00 BSC			
Exposed Pad Width	E2		4.00		
Overall Length	D		6.00 BSC		
Exposed Pad Length	D2		4.00		
Terminal Width	b	0.35	0.40	0.45	
Corner Pad	b1	0.55	0.60	0.65	
Corner Pad, Metal Free Zone	b2	0.15	0.20	0.25	
Terminal Length	L	0.55	0.60	0.65	
Terminal-to-Exposed Pad	K	0.20	-	-	

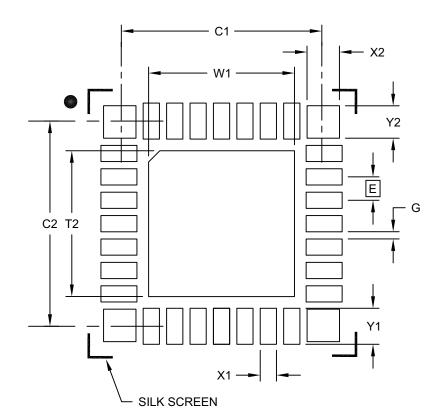
Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. Package is saw singulated
- 3. Dimensioning and tolerancing per ASME Y14.5M
 - BSC: Basic Dimension. Theoretically exact value shown without tolerances. REF: Reference Dimension, usually without tolerance, for information purposes only.
- 4. Outermost portions of corner structures may vary slightly.

Microchip Technology Drawing C04-0209 Rev C Sheet 2 of 2

28-Lead Plastic Quad Flat, No Lead Package (MX) - 6x6 mm Body [UQFN] With 0.60mm Contact Length And Corner Anchors

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

Units		Ν	MILLIMETER	S
Dimensio	n Limits	MIN	NOM	MAX
Contact Pitch	E		0.65 BSC	
Optional Center Pad Width	W1			4.05
Optional Center Pad Length	T2			4.05
Contact Pad Spacing	C1		5.70	
Contact Pad Spacing	C2		5.70	
Contact Pad Width (X28)	X1			0.45
Contact Pad Length (X28)	Y1			1.00
Corner Pad Width (X4)	X2			0.90
Corner Pad Length (X4)	Y2			0.90
Distance Between Pads	G	0.20		

Notes:

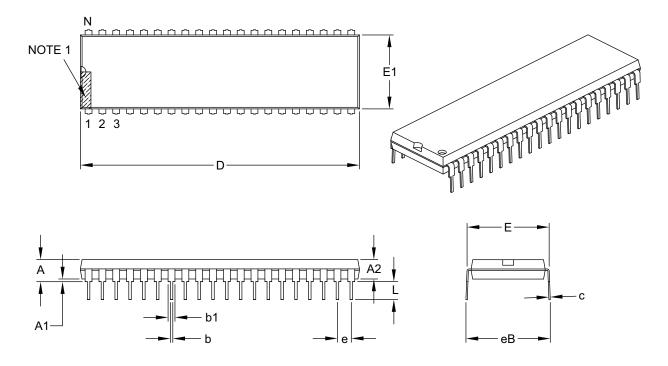
1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2209B

40-Lead Plastic Dual In-Line (P) – 600 mil Body [PDIP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units		INCHES	
Dimension	n Limits	MIN	NOM	MAX
Number of Pins	Ν		40	
Pitch	е		.100 BSC	
Top to Seating Plane	Α	-	-	.250
Molded Package Thickness	A2	.125	-	.195
Base to Seating Plane	A1	.015	-	-
Shoulder to Shoulder Width	E	.590	-	.625
Molded Package Width	E1	.485	-	.580
Overall Length	D	1.980	-	2.095
Tip to Seating Plane	L	.115	-	.200
Lead Thickness	С	.008	-	.015
Upper Lead Width	b1	.030	-	.070
Lower Lead Width	b	.014	-	.023
Overall Row Spacing §	eВ	-	_	.700

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

2. § Significant Characteristic.

3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" per side.

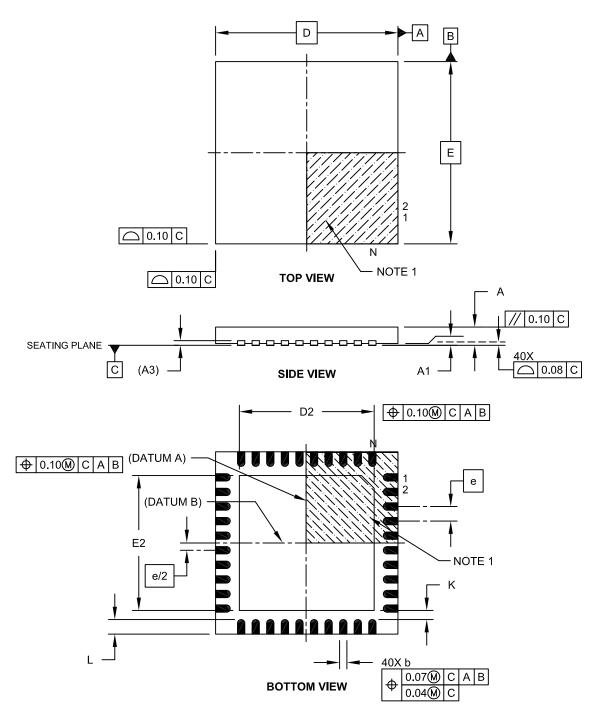
4. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-016B

40-Lead Ultra Thin Plastic Quad Flat, No Lead Package (MV) – 5x5x0.5 mm Body [UQFN]

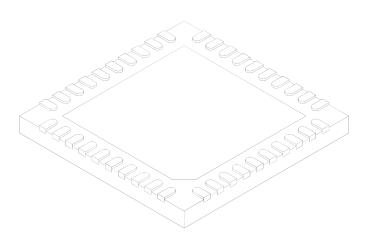
Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Microchip Technology Drawing C04-156A Sheet 1 of 2

40-Lead Ultra Thin Plastic Quad Flat, No Lead Package (MV) – 5x5x0.5 mm Body [UQFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units	Ν	/ILLIMETER	S
Dimensio	on Limits	MIN	NOM	MAX
Number of Pins	N		40	
Pitch	е		0.40 BSC	
Overall Height	A	0.45	0.50	0.55
Standoff	A1	0.00	0.02	0.05
Contact Thickness	A3	0.127 REF		
Overall Width	E	5.00 BSC		
Exposed Pad Width	E2	3.60	3.70	3.80
Overall Length	D	5.00 BSC		
Exposed Pad Length	D2	3.60	3.70	3.80
Contact Width	b	0.15	0.20	0.25
Contact Length	L	0.30	0.40	0.50
Contact-to-Exposed Pad	K	0.20	-	-

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

2. Package is saw singulated.

3. Dimensioning and tolerancing per ASME Y14.5M.

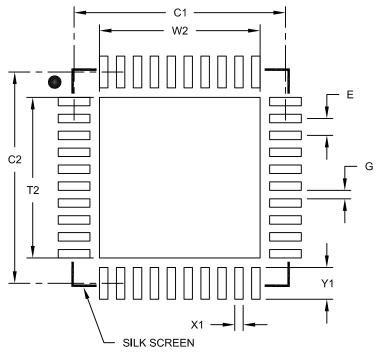
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-156A Sheet 2 of 2

40-Lead Plastic Ultra Thin Quad Flat, No Lead Package (MV) - 5x5 mm Body [UQFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

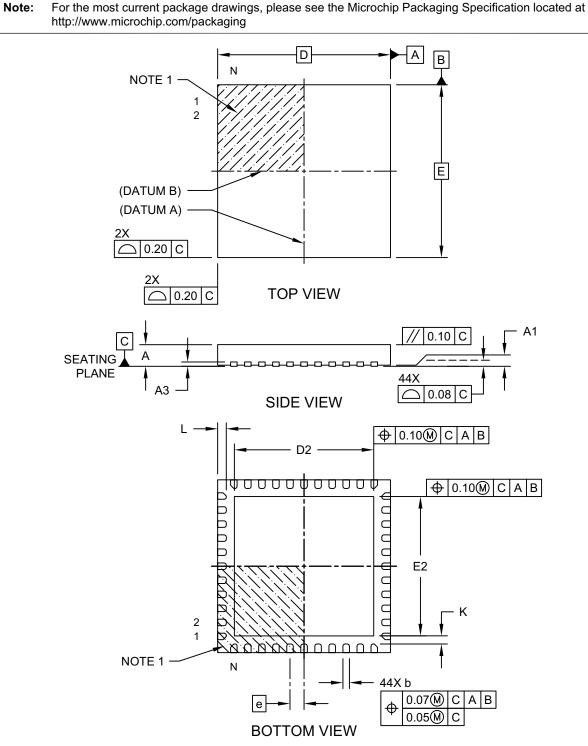
Units		MILLIMETERS		
Dimensior	Dimension Limits		NOM	MAX
Contact Pitch	E	E 0.40 BS		•
Optional Center Pad Width	W2			3.80
Optional Center Pad Length	T2			3.80
Contact Pad Spacing	C1		5.00	
Contact Pad Spacing	C2		5.00	
Contact Pad Width (X40)	X1			0.20
Contact Pad Length (X40)	Y1			0.75
Distance Between Pads	G	0.20		

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2156B

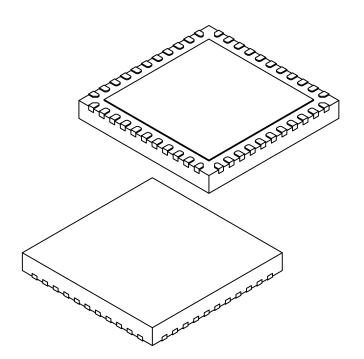


44-Lead Plastic Quad Flat, No Lead Package (ML) - 8x8 mm Body [QFN or VQFN]

Microchip Technology Drawing C04-103D Sheet 1 of 2

44-Lead Plastic Quad Flat, No Lead Package (ML) - 8x8 mm Body [QFN or VQFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		MILLIMETERS		
Dimension	Limits	MIN	NOM	MAX
Number of Pins	N		44	
Pitch	е		0.65 BSC	
Overall Height	Α	0.80	0.90	1.00
Standoff	A1	0.00	0.02	0.05
Terminal Thickness	A3	0.20 REF		
Overall Width	E	8.00 BSC		
Exposed Pad Width	E2	6.25	6.45	6.60
Overall Length	D		8.00 BSC	
Exposed Pad Length	D2	6.25	6.45	6.60
Terminal Width	b	0.20	0.30	0.35
Terminal Length	L	0.30	0.40	0.50
Terminal-to-Exposed-Pad	К	0.20	-	-

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

- 2. Package is saw singulated
- 3. Dimensioning and tolerancing per ASME Y14.5M

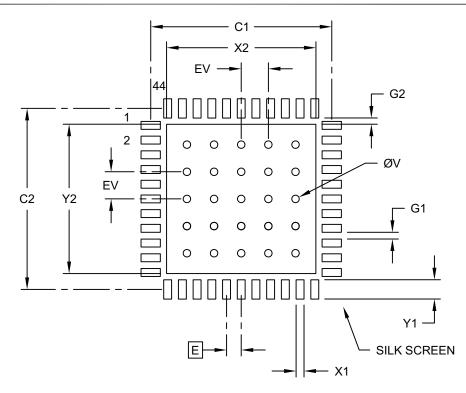
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-103D Sheet 2 of 2

44-Lead Plastic Quad Flat, No Lead Package (ML) - 8x8 mm Body [QFN or VQFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

Units		Ν	IILLIMETER	S
Dimension	Limits	MIN	NOM	MAX
Contact Pitch	E		0.65 BSC	
Optional Center Pad Width	X2			6.60
Optional Center Pad Length	Y2			6.60
Contact Pad Spacing	C1		8.00	
Contact Pad Spacing	C2		8.00	
Contact Pad Width (X44)	X1			0.35
Contact Pad Length (X44)	Y1			0.85
Contact Pad to Contact Pad (X40)	G1	0.30		
Contact Pad to Center Pad (X44)	G2	0.28		
Thermal Via Diameter	V		0.33	
Thermal Via Pitch	EV		1.20	

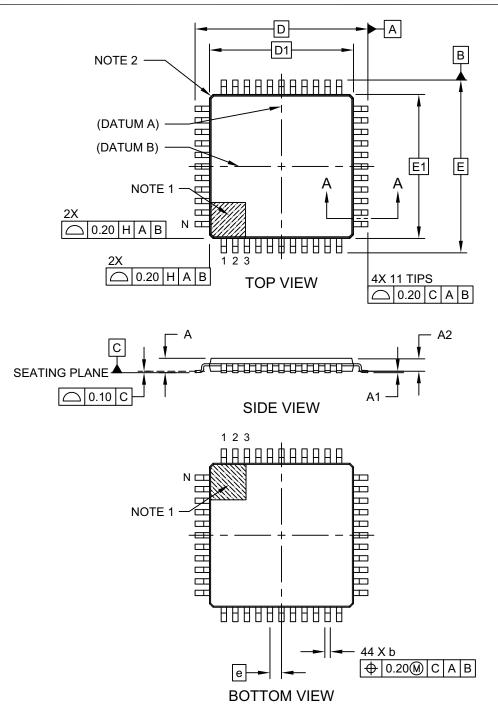
Notes:

- 1. Dimensioning and tolerancing per ASME Y14.5M
 - BSC: Basic Dimension. Theoretically exact value shown without tolerances.
- 2. For best soldering results, thermal vias, if used, should be filled or tented to avoid solder loss during reflow process

Microchip Technology Drawing No. C04-2103C

44-Lead Plastic Thin Quad Flatpack (PT) - 10x10x1.0 mm Body [TQFP]

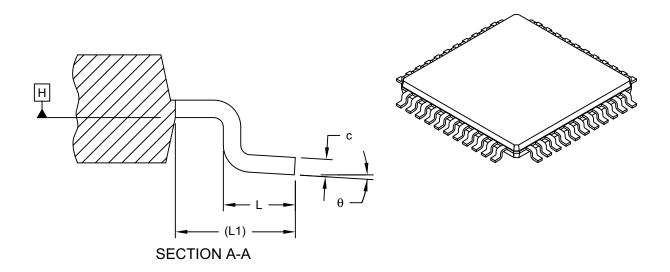
Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Microchip Technology Drawing C04-076C Sheet 1 of 2

44-Lead Plastic Thin Quad Flatpack (PT) - 10x10x1.0 mm Body [TQFP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		Ν	ILLIMETER	S
Dimension	Limits	MIN	NOM	MAX
Number of Leads	N		44	
Lead Pitch	е		0.80 BSC	
Overall Height	Α	-	-	1.20
Standoff	A1	0.05	-	0.15
Molded Package Thickness	A2	0.95	1.00	1.05
Overall Width	E	12.00 BSC		
Molded Package Width	E1	10.00 BSC		
Overall Length	D		12.00 BSC	
Molded Package Length	D1		10.00 BSC	
Lead Width	b	0.30	0.37	0.45
Lead Thickness	С	0.09	-	0.20
Lead Length	L	0.45	0.60	0.75
Footprint	L1	1.00 REF		
Foot Angle	θ	0°	3.5°	7°

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

2. Exact shape of each corner is optional.

3. Dimensioning and tolerancing per ASME Y14.5M

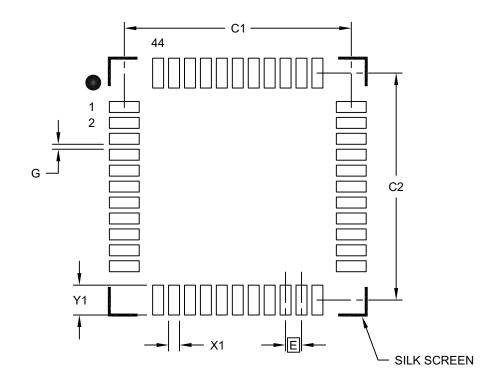
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-076C Sheet 2 of 2

44-Lead Plastic Thin Quad Flatpack (PT) - 10X10X1 mm Body, 2.00 mm Footprint [TQFP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

Units		Ν	ILLIMETER	S
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E		0.80 BSC	
Contact Pad Spacing	C1		11.40	
Contact Pad Spacing	C2		11.40	
Contact Pad Width (X44)	X1			0.55
Contact Pad Length (X44)	Y1			1.50
Distance Between Pads	G	0.25		

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2076B

APPENDIX A: DATA SHEET REVISION HISTORY

Revision A (11/2015)

Initial release of this document.

Revision B (10/2016)

Updated Figures 14-1, 23-3, 23-8, 23-9, and 23-10; Registers 7-5, 7-11, 18-1, 19-1, 24-6, 27-11, 31-3, 31-4, 31-5, 31-6, 31-7, and 32-4; Section 32.6; Tables 3, 4, 3-4, 3-6, 3-7, 3-14, 3-15, 3-18, 12-1, 12-2, 12-3, 24-4, 25-5, 27-5, 27-6, 28-1, 32-4, 36-1, 36-2, 36-7 and 36-8.

Updated the Cover page.

Section 20.5 rewritten. Added Characterization Data.

THE MICROCHIP WEBSITE

Microchip provides online support via our website at www.microchip.com. This website is used as a means to make files and information easily available to customers. Accessible by using your favorite Internet browser, the website contains the following information:

- Product Support Data sheets and errata, application notes and sample programs, design resources, user's guides and hardware support documents, latest software releases and archived software
- General Technical Support Frequently Asked Questions (FAQ), technical support requests, online discussion groups, Microchip consultant program member listing
- Business of Microchip Product selector and ordering guides, latest Microchip press releases, listing of seminars and events, listings of Microchip sales offices, distributors and factory representatives

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- Field Application Engineer (FAE)
- Technical Support

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Technical support is available through the website at: http://www.microchip.com/support

PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

PART NO.	[X] ⁽¹⁾ - X /XX XXX ☐ ☐ ☐ Tape and Reel Temperature Package Patterr Option Range	Industrial temperature PDIP package
Device:	PIC16F1777, PIC16LF1777, PIC16F1778, PIC16LF1778, PIC16F1779, PIC16F1779	b) PIC16F1779-E/SS Extended temperature SSOP package
Tape and Reel Option:	Blank = Standard packaging (tube or tray) T = Tape and Reel ⁽¹⁾	
Temperature Range:	I = -40° C to $+85^{\circ}$ C (Industrial) E = -40° C to $+125^{\circ}$ C (Extended)	
Package: ⁽²⁾	$\begin{array}{llllllllllllllllllllllllllllllllllll$	 Note 1: Tape and Reel identifier only appears in the catalog part number description. This identifier is used for ordering purposes and is not printed on the device package. Check with your Microchip Sales Office for package availability with the Tape and Reel option. 2: Small form-factor packaging options may be available. Places check
Pattern:	QTP, SQTP, Code or Special Requirements (blank otherwise)	be available. Please check www.microchip.com/packaging for small-form factor package availability, or contact your local Sales Office.

PIC16(L)F1777/8/9

NOTES:

Note the following details of the code protection feature on Microchip devices:

- · Microchip products meet the specification contained in their particular Microchip Data Sheet.
- Microchip believes that its family of products is one of the most secure families of its kind on the market today, when used in the intended manner and under normal conditions.
- There are dishonest and possibly illegal methods used to breach the code protection feature. All of these methods, to our knowledge, require using the Microchip products in a manner outside the operating specifications contained in Microchip's Data Sheets. Most likely, the person doing so is engaged in theft of intellectual property.
- Microchip is willing to work with the customer who is concerned about the integrity of their code.
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