Bi-CMOS IC

For Portable Electronic Device Use 1.4W × 2ch BTL Power Amplifier



http://onsemi.com

Overview

The LV49821VH incorporates a 2-channel power circuit amplifier capable of low-voltage operation (2.7V and up). It has a function for switching the headphone driver and also has a standby function to reduce the current drain. It is a power amplifier IC optimal for driving the speakers used in portable equipment and low power output equipment.

Use

• Portable DVD-player, Note PC, Portable TV, LCD monitor, Active speaker, and more.

Features

- 2-cannels BTL power amplifier built-in: Standard output power = 1.4W (V_{CC} = 5V, R_L = 8 Ω , THD = 10%) Output coupling capacitor is unnecessary because of differential output type.
- Standby function built-in: Standard standby current = $0.01\mu A$ (V_{CC} = 5V)
- Second amplifier stop control function built-in: Headphone driver switch (for BTL/SE switch)

 Audio mute (Only BTL power amplifier path)
- Supports beep signal input
- Thermal protection circuit built-in
- Operation at low voltage possible: $V_{CC} = 2.7V$ to 5.5V
- Gain setting possible: BTL voltage gain = 0 to 26dB

Specifications

Maximum Ratings at $Ta = 25^{\circ}C$

Parameter	Symbol	Conditions	Ratings	Unit
Maximum supply voltage	VCC max		6	٧
Allowable power dissipation	Pd max	Mounted on a specified board.*	1.5	W
Maximum junction temperature	Tj max		150	°C
Operating temperature	Topr		-30 to +75	°C
Strage temperature	Tstg		-40 to +150	℃

Specified board (Our company Evaluation board): 70mm × 70mm × 1.6mm, glass epoxy both side.

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

Operating Conditions at $Ta = 25^{\circ}C$

Parameter	Symbol	Conditions	Ratings	Unit
Recommended supply voltage	VCC		5	>
Recommended load resistance	RL		4 to 32	Ω
Operation supply voltage range	V _{CC} op1	$R_L = 8\Omega$ or more	2.7 to 5.5	V
	V _{CC} op2	$R_L = 4\Omega$ or 6Ω	2.7 to 3.7	V

Note: Please determine supply voltage used with due consideration of allowable power dissipation.

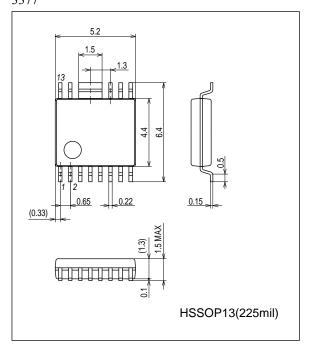
Electrical Characteristics at $Ta = 25^{\circ}C$, $V_{CC} = 5V$, fin = 1 kHz, $R_L = 8\Omega$, V2 = high, V6 = Low

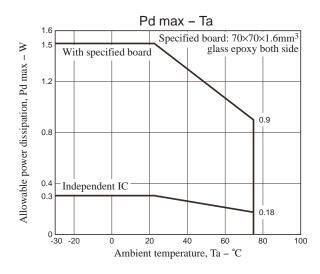
Danamatan	O. made ad	Ourselved.	Ratings			11.2
Parameter	Symbol Conditions		min	typ	max	Unit
Quiescent current	ICCOP1	No signal, R _L = ∞		7.1	13	mA
	I _{CCO} P2	No signal, R _L = ∞,V6 = High (2nd amplifier stop)		4.0		mA
Standby current	ISTBY	No signal, R _L = ∞,V2 = Low (Standby mode)		0.01	10	μΑ
Maximum output power	P _O max	THD = 10%	0.91	1.4		W
Voltage gain	V _G	Vin = -25dBV	17.6	19.1	20.6	dB
Voltage gain difference	V _{GR}		0		26	dB
Channel balance	СНВ	Vin = -25dBV	-1.5	0	+1.5	dB
Total harmonic distortion	THD	Vin = -25dBV		0.3	1	%
Output noise voltage	V _{NO}	Rg = 620Ω , 20 to $20kHz$		35	100	μVrms
Channel separation	CHsep	Vout = -25dBv, 20 to 20kHz	50	60		dB
Output offset voltage	V _{DC} OS	$Rg = 620\Omega$	-30		+30	mV
Muting attenuation level	MUTE1	Vin = 0dBV, V2 = Low (Standby mode)		-110		dBV
	MUTE2	Vin = -10dBV,V6 = High (2nd amplifier stop)		-85		dBV
Ripple rejection ratio	SVRR	Rg = 620Ω , fr = 100 Hz, Vr = -20 dBV		35		dB
Reference Voltage (pin 4)	Vref			2.5		V
High level control voltage (pin 2)	V _{STB} H	Power amplifier operation mode	1.6		Vcc	V
Low level control voltage (pin 2)	V _{STB} L	Power amplifier standby mode	0		0.3	V
High level control voltage (pin 6)	V _{2CNT} H	2nd amplifier standby mode (SE mode)	4V _{CC} /5	_	Vcc	V
Low level control voltage (pin 6)	V _{2CNT} H	2nd amplifier operation mode (BTL mode)	0		V _{CC} /2	V

Package Dimensions

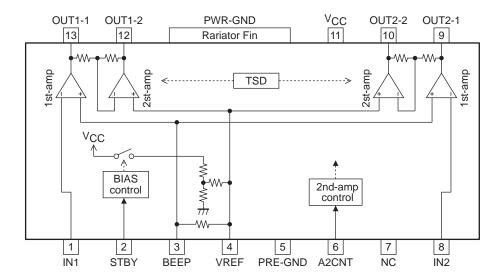
unit: mm (typ)

3377

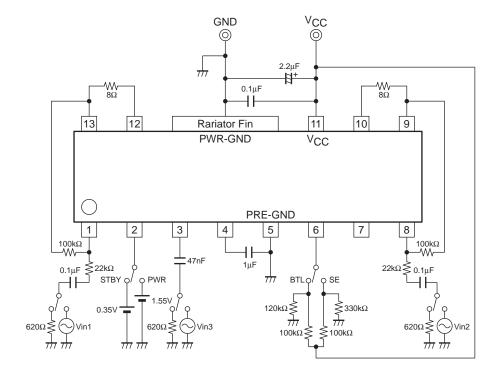




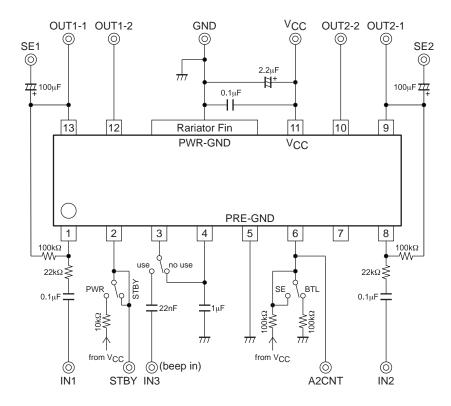
Block Diagram



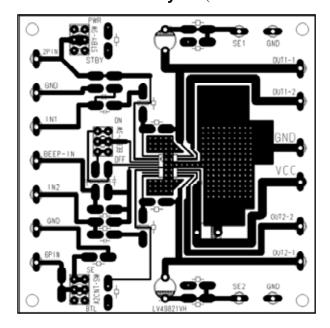
Test Circuit

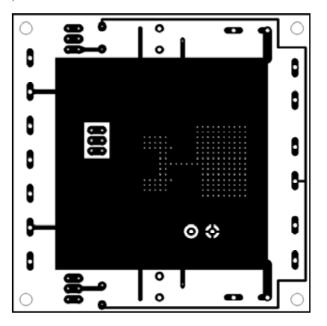


Evaluation Board Circuit



Evaluation Board Layout (70mm × 70mm × 1.6mm)



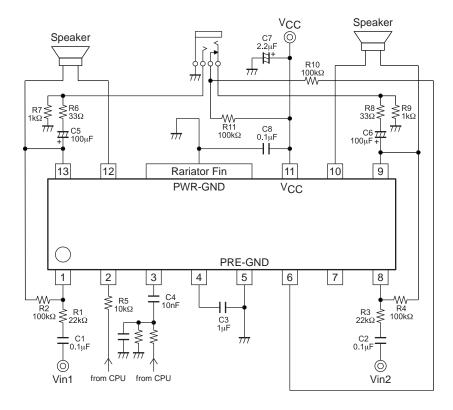


Top layer

Bottom Layer

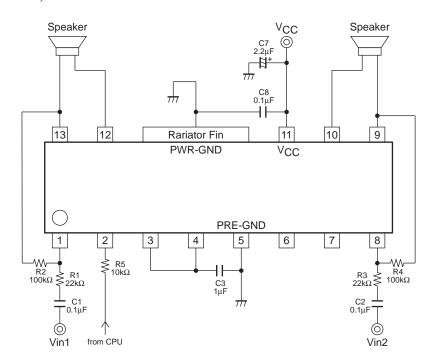
Application Circuit Example 1

(BTL/SE switching function use)



Application Circuit Example 2

(Only BTL function use)



Pin Function

	unction			
Pin No.	Pin name	Pin voltage V _{CC} =5V	Function	Equivalent circuit
1 8	IN1 IN2	2.5V 2.5V	Power amplifier input pin (1ch). Power amplifier input pin (2ch).	VCC VCC VCC VREF2 GND
2	STBY	External apply	Standby control pin. • Standby mode (0 to 0.3V) • Operation mode (1.6V to V _{CC})	VCC 30kΩ BIAS GND
3 4	BEEP VREF	2.5V 2.5V	Beep signal input pin. Reference voltage pin.	VCC VREF2 VREF VCC VREF2 VREF2 VCC VREF2
5	PRE-GND	0V	Pre-stage block ground pin	
6	A2CNT	External apply	2nd amplifier stop control pin.	6 VCC GAND GND
7	NC	_	Unused pin.	
9 13	OUT2-1 OUT1-1	2.5V 2.5V	BTL 1st output pin (2ch). BTL 1st output pin (1ch).	VCC VREF2 VCC VREF2 9 13 9 13
10 12	OUT2-2 OUT1-2	2.5V 2.5V	BTL 2nd output pin (2ch). BTL 2nd output pin (1ch).	VCC VREF (10)12
11	V _{CC}	External apply	Power supply pin.	
FIN	PWR-GND	0V	Power system ground pin, Radiation fin.	

Usage Note

1. Input coupling capacitor (C1 and C2)

C1 and C2 are input coupling capacitors that are used to cut the DC component. The input coupling capacitors C1, C2 and the input resistors R1 and R3 make up the high-pass filter, attenuating the bass frequency. Therefore, the capacitance value must be selected with due consideration of the cut-off frequency.

The cut-off frequencies are expressed by the following formulas.

1 ch
$$\Rightarrow$$
 fc1 = 1/(2 π × C1 × R1)
2 ch \Rightarrow fc2 = 1/(2 π × C2 × R3)

This capacitor affects the pop sound at startup. Note with care that increasing the capacitance value lengthens the charging time of the capacitor, which will make the pop sound louder.

2. BTL voltage gain

The voltage gain of the first amplifier is determined by the ratio between the resistors R1 and R2 (R3 and R4).

$$1 \text{ch} \Rightarrow \text{Vg1} = 20 \times \log(\text{R2/R1}) \dots \text{unit: dB}$$

$$2ch \Rightarrow Vg2 = 20 \times log(R4/R3) \dots unit: dB$$

Therefore, the BTL voltage gain is expressed by the following formulas.

1ch
$$\Rightarrow$$
 VgBTL1 = 6 + 20 \times log(R2/R1) ... unit: dB

$$2ch \Rightarrow VgBTL2 = 6 + 20 \times log(R4/R3) \dots unit: dB$$

The BTL voltage gain must be set in the range of 0 to 26 dB.

3. Beep signal input pin (pin 3)

This pin is connected to the non-inverting input block of the first amplifier of the BTL amplifier, and is biased internally by a $50k\Omega$ resistor. The input coupling capacitor C4 and the bias resistor make up a high-pass filter that attenuates bass band signals, so when determining the C4 capacitance value, the value must be set with due consideration of the cut-off frequency. The cut-off frequency is expressed by the following formula.

$$fc3 = 1/(2\pi \times C4 \times 50000)$$

In addition, when input from Pin 3, the BTL voltage gain is expressed by the following formulas.

$$1ch \Rightarrow VgBTL1 = 6 + 20 \times log(1+R2/(R1 + ro)) \dots unit: dB$$

$$2ch \Rightarrow VgBTL2 = 6 + 20 \times log(1+R4/(R3 + ro)) \dots unit: dB$$

When setting the signal level, the signal should be attenuated and input as shown in Fig.1.

When not using this input pin, connect it to pin 4 as shown in Application Circuit Example-2.

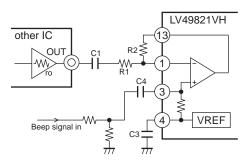


Fig.1

4. pin 4 capacitor (C3)

This capacitor is a ripple filter capacitor. The internal resistors $(100k\Omega + 450k\Omega)$ and C3 make up a low-pass filter that is used to reduce the power supply ripple component and increase the ripple rejection ratio.

Note that inside the IC, the rising-transient-response-characteristic of the pin 4 voltage (reference voltage) is used to activate the automatic pop sound reduction circuit. Therefore, when reducing the C3 capacitance value to increase the voltage rise speed, the design should take into account that the pop sound increases during voltage rise.

5. Power supply line capacitor (C7and C8)

The bypass capacitor C8 is used to remove the high frequency component that cannot be eliminated by the power supply capacitor C7 (chemical capacitor). Place the bypass capacitor C8 as near to the IC as possible, and use a ceramic capacitor with good high frequency characteristics.

When using a stabilized power supply, these capacitors can also be combined into a single 2.2μ F ceramic capacitor. Note that when the power supply line is relatively unstable, the power supply capacitor C7 capacitance value must be increased.

6. Standby pin (pin 2)

By controlling the standby pin, the mode changeover can be made between standby and operation modes. Direct control is possible using the CPU output port, but inserting a series resistor R5 (1 k Ω or more) is recommended in case the pin is affected by digital noise from the CPU.

Standby mode ...
$$V2 = 0V$$
 to $0.3V$

Operating mode...
$$V2 = 1.6V$$
 to V_{CC}

In addition, when not using standby mode, this pin can also be used interlocked with the power supply as shown in Fig. 2. The series resistor R5 can be eliminated, but the current I2 expressed by the following formula flows through the standby pin, so this should be taken into account in the design.

Pin 2 inflow current (unit: A):
$$I2 = 7 \times 10^{-6} + (V_{CC} - 0.7) / (R5 + 30000)$$

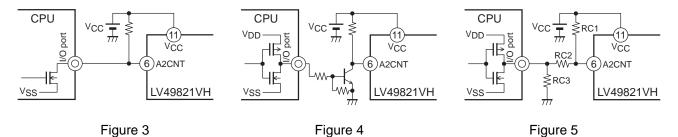
Fig.2

7. Pin 6 control (2nd amplifier stop control function)

Pin 6 performs on/off control for the BTL amplifier's second amplifier operation. This function enables switching between speaker drive (BTL output system) and headphone drive (single end output system). The control comparator is connected to this pin, and this threshold voltage is generated by resistance division from the supply voltage. For this reason, care should be taken, as the threshold value varies according to the supply voltage. When switching using a headphone jack switch, the connection method shown in Application Circuit Example-1 is recommended.

Comparator threshold value: Vth =
$$V_{CC} \times 2/3$$

In addition, when controlling this pin with the CPU (BTL amplifier mute function), care should be taken for the relationship between the supply voltage used by the CPU and the supply voltage used by the power amplifier IC. When the supply voltage used by the power amplifier IC is higher, open/low format control as shown in Fig.3 and Fig.4 is recommended. In addition, there is also a control method that uses three resistors as shown in Fig.5. The recommended ratio between the resistance values of these three resistors is as follows.



8. Headphone drive

When also using the BTL amplifier's first amplifier as the headphone amplifier, it is recommended to adjust the level by inserting series resistors R6 and R8 to the signal line as shown in Application Circuit Example-1.

Note that this series resistor, the headphone load resistance and the output coupling capacitors C5 and C6 make up a high-pass filter, so this should be taken into account in the design. The cut-off frequencies are expressed by the following formulas.

1ch
$$\Rightarrow$$
 fc1 = 1 / $(2\pi \times C5 \times (R6 + R_L))$
2ch \Rightarrow fc2 = 1 / $(2\pi \times C6 \times (R8 + R_L))$

9. Load capacitance

When connecting a capacitor between the output pin and ground to suppress electromagnetic radiation or other purposes, the effects of this capacitor may cause the power amplifier phase margin to be reduced, resulting in oscillation. When adding this capacitor, care should be taken for the capacitance value.

Recommended capacitance value: 0.033µF to 0.33µF

10. Thermal protection circuit

The IC has a built-in thermal protection circuit that can reduce the risk of breakdown or degradation when the IC becomes abnormally hot for some reason. When the internal chip junction temperature Tj rises to approximately 170° C, this protective circuit operates to cut off the power supply to the power amplifier block and stop signal output. Operation recovers automatically when the chip temperature drops to approximately 130°C.

Note that this circuit cannot always prevent breakdown or degradation, so sufficient care should be taken for using the IC. When the chip becomes abnormally hot, immediately turn off the power and determine the cause.

11. Short-circuit between pins

Turning on the power supply with the short-circuit between terminals leads to the deterioration and destruction of IC. When fixing the IC to the substrate, please check that the solder is not short-circuited between the terminals before turning on the power.

12. Load Short-circuit

Leaving the IC in the load short-circuit for many hours leads to the deterioration and destruction of the IC. The load must not be short-circuited absolutely.

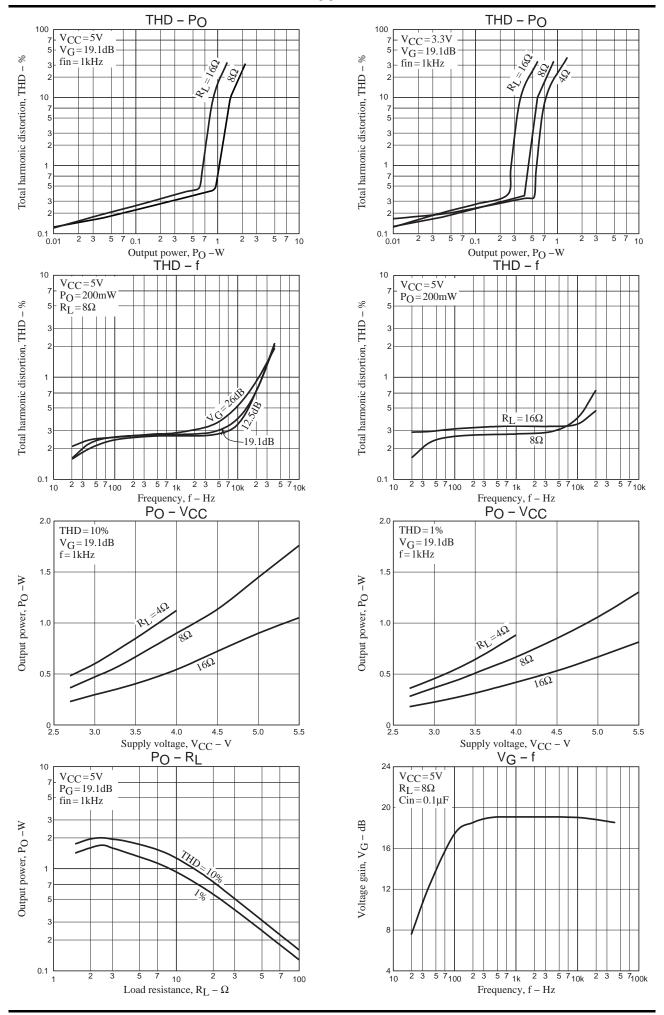
13. Maximum rating

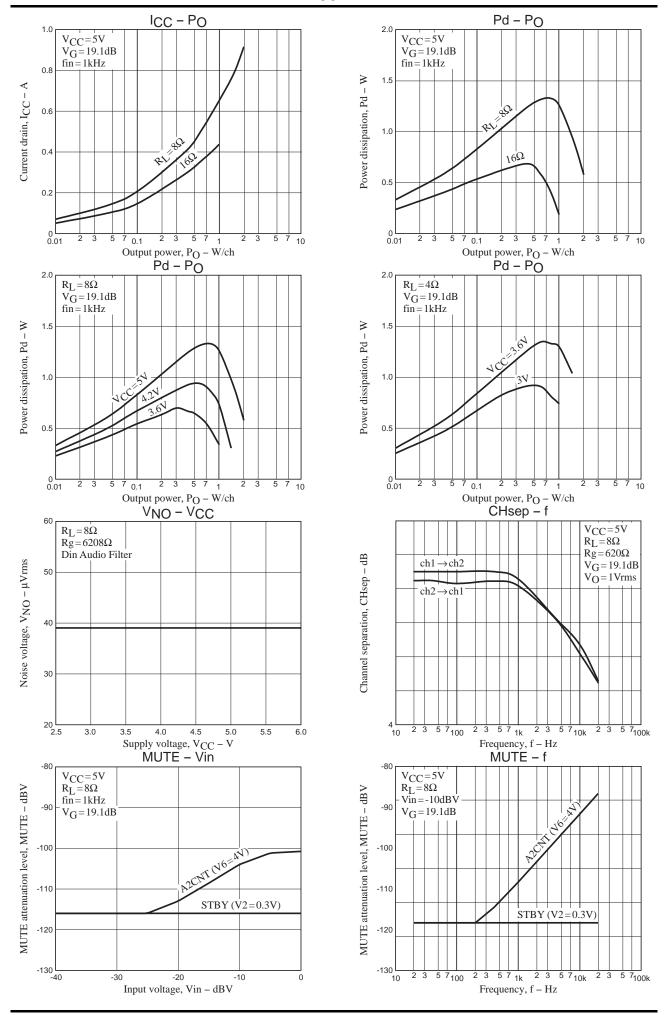
When the rated value used is just below to the absolute maximum ratings value, there is a possibility to exceed the maximum rating value with slight extrusion variable. Also, it can be a destructive accident.

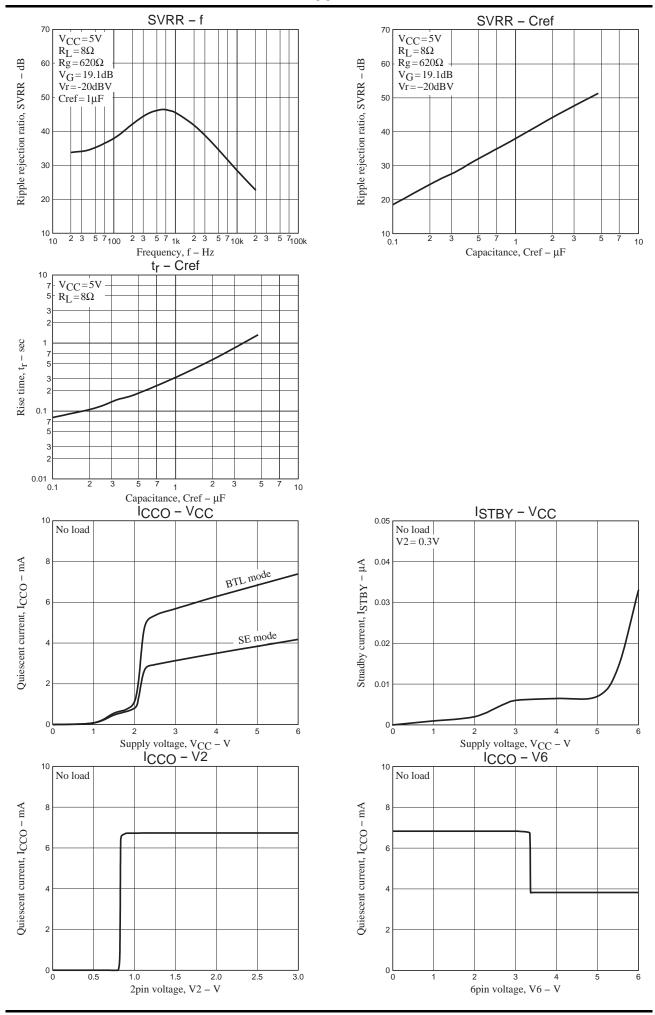
Please use within the absolute maximum ratings with sufficient variation margin of supply voltage.

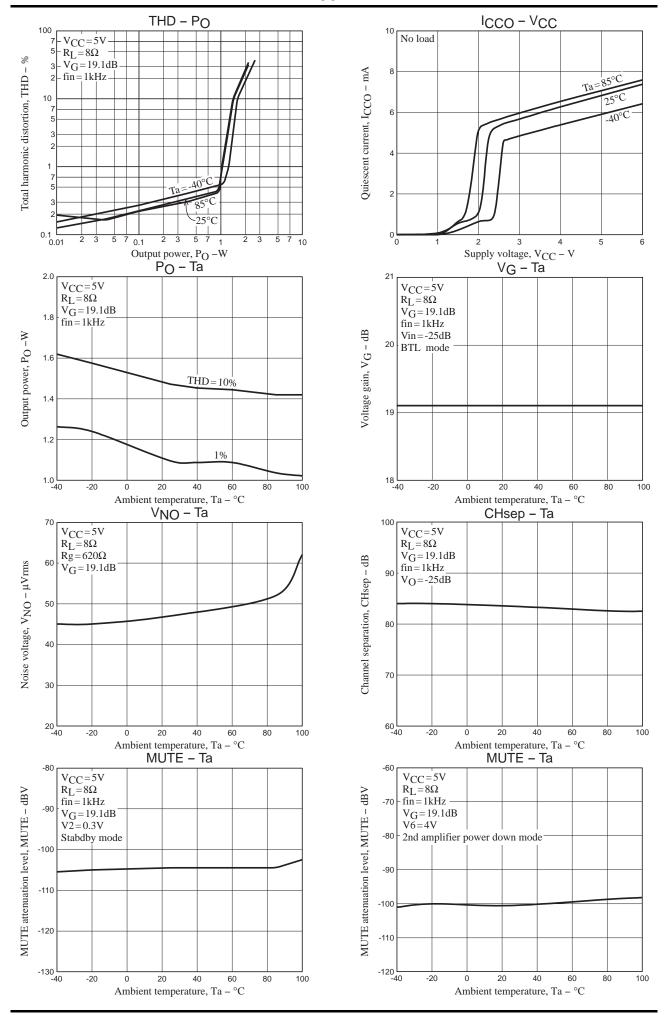
In addition, the package of this IC has low thermal radiation characteristics, so secure sufficient thermal radiation by providing a copper foil land on the printed circuit board near the heat sink.

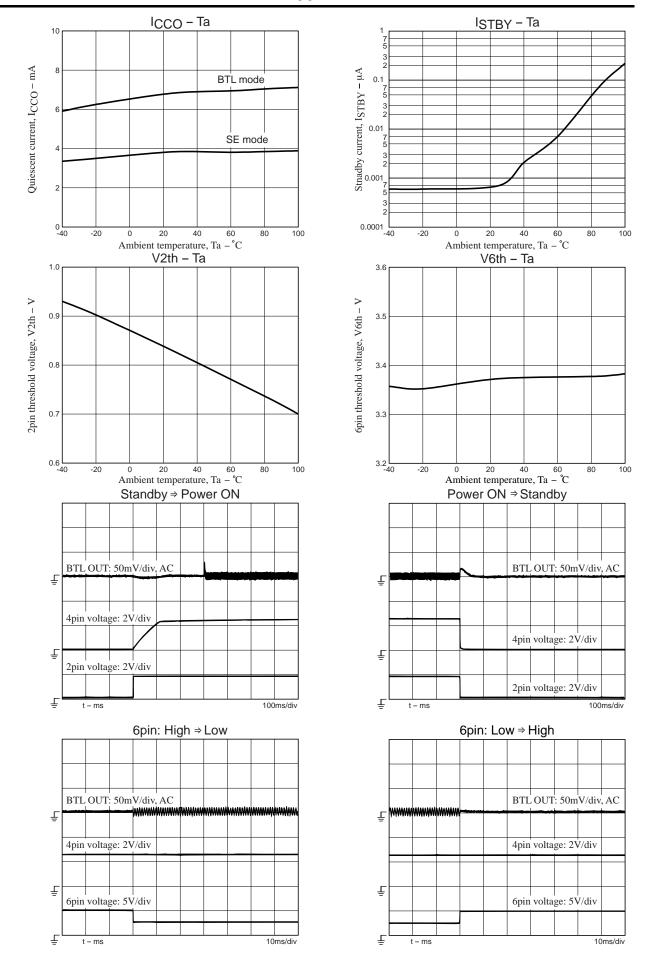
When $V_{CC} = 5V$ and load = 8Ω , a ground line copper foil area of approximately $50 \text{mm} \times 50 \text{mm}$ is recommended.











ON Semiconductor and the ON logo are registered trademarks of Semiconductor Components Industries, LLC (SCILLC). SCILLC owns the rights to a number of patents, trademarks, copyrights, trade secrets, and other intellectual property. A listing of SCILLC's product/patent coverage may be accessed at www.onsemi.com/site/pdf/Patent-Marking.pdf. SCILLC reserves the right to make changes without further notice to any products herein. SCILLC makes no warranty, representation or guarantee regarding the suitability of its products for any particular purpose, nor does SCILLC assume any liability arising out of the application or use of any product or circuit, and specifically disclaims any and all liability, including without limitation special, consequential or incidental damages. "Typical" parameters which may be provided in SCILLC data sheets and/or specifications can and do vary in different applications and actual performance may vary over time. All operating parameters, including "Typicals" must be validated for each customer application by customer's technical experts. SCILLC does not convey any license under its patent rights nor the rights of others. SCILLC products are not designed, intended, or authorized for use as components in systems intended for surgical implant into the body, or other applications intended to support or sustain life, or for any other application in which the failure of the SCILLC product could create a situation where personal injury or death may occur. Should Buyer purchase or use SCILLC products for any such unintended or unauthorized application, Buyer shall indemnify and hold SCILLC and its officers, employees, subsidiaries, affiliates, and distributors harmless against all claims, costs, damages, and expenses, and reasonable attorney fees arising out of, directly or indirectly, any claim of personal injury or death associated with such unintended or unauthorized use, even if such claim alleges that SCILLC was negligent regarding the design or manufacture of the part. SCILLC is an Equa

X-ON Electronics

Largest Supplier of Electrical and Electronic Components

Click to view similar products for Audio Amplifiers category:

Click to view products by ON Semiconductor manufacturer:

Other Similar products are found below:

LV47002P-E MP7747DQ-LF-P AZ386MTR-E1 NCP2811AFCT1G NCP2890AFCT2G NJM8068RB1-TE1 NJW1194V-TE1 LA4282-E
LA4814JA-AE LC706200CM SSM2377ACBZ-R7 FDA2100LV TDA2541 TDA7385H TDA7391LV TDA7575BPDTR TDA7718NTR

IS31AP2121-LQLS1 IS31AP4915A-QFLS2-TR LA74309FA-BH 421067X 480263C NCP2820FCT2G STPA001 TDA1515AQ TDA1520B

TDA1591T TDA2051H TDA4850 TDA7391PDUTR TDA7563BH TDA7718B LA4425F-E LA4742-E TDA7391PDU

TDA7491MV13TR TDA749213TR TDA7563AH TDA7850H STK433-070GN-E E-TDA7391PDTR SSM2529ACBZ-R7 SSM2518CBZ
R7 MAX9890BEBL+T MAX98303EWE+T MAX98358EWL+ MAX98304DEWL+T MAX97220DETE+T TS4962MEIJT TS4990EIJT