



MIC7111

1.8V to 11V, 15 μ A, 25kHz GBW, Rail-to-Rail
Input and Output Operational Amplifier

General Description

The MIC7111 is a low-power operational amplifier with rail-to-rail inputs and outputs. The device operates from a 1.8V to 11V single supply or an $\pm 0.9V$ to $\pm 5.5V$ dual supply. The device consumes a low 15 μ A of current from a 1.8V supply and 25 μ A from a 10V supply. The device features a unity gain bandwidth of 25kHz and swings within 1mV of either the supply rail with a 100k Ω load. The device is capable of sinking and sourcing 25mA of current from a 1.8V supply and up to 200mA from a 10V supply. The device is available in the cost effective SOT23-5 package.

Datasheets and support documentation are available on Micrel's web site at: www.micrel.com.

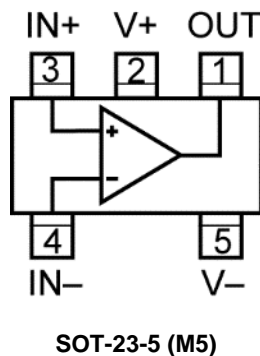
Features

- 1.8V to 11V single supply operation
- $\pm 0.9V$ to $\pm 5.5V$ dual supply operation
- Low 15 μ A supply current at 1.8V
- 25kHz gain bandwidth
- 1mV input offset voltage (typical)
- 1pA input bias current (typical)
- 0.01pA input offset current (typical)
- Input-referred noise is 110nv/ $\sqrt{\text{Hz}}$ at 1kHz
- Output swing to within 1mV of rails with 1.8V supply and 100k Ω load
- Suitable for driving capacitive loads
- Cost effective SOT23-5 package

Applications

- Wireless and cellular communications
- GaAs RF bias amplifier
- Current sensing for battery chargers
- Transducer linearization and interface
- Portable computing

Functional Configuration



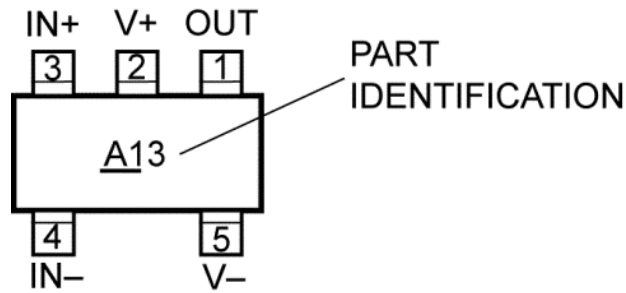
Ordering Information

Part Number	Junction Temperature Range	Package ⁽¹⁾
Pb-Free		
MIC7111YM5	-40°C to +85°C	SOT23-5

Note:

1. Other packages are available. Contact Micrel for details.

Pin Configuration



SOT23-5 (M5)
(Top View)

Pin Description

Pin Number	Pin Name	Pin Function
1	OUT	Amplifier Output.
2	V+	Positive Supply
3	IN+	Non-inverting Input.
4	IN-	Inverting Input
5	V-	Negative Supply.

Absolute Maximum Ratings⁽¹⁾

Supply Voltage ($V_{V+} - V_{V-}$)	+12V
Differential Input Voltage ($V_{IN+} - V_{IN-}$)	$\pm(V_{V+} - V_{V-})$
I/O Pin Voltage (V_{IN}, V_{OUT}) ⁽³⁾	$V_{V+} + 0.3V$ to $V_{V-} - 0.3V$
Junction Temperature (T_J)	+150°C
Lead Temperature (soldering, 10s)	260°C
Storage Temperature (T_s)	-65°C to +150°C
ESD Rating ⁽⁶⁾	2kV

Operating Ratings⁽²⁾

Supply Voltage ($V_{V+} - V_{V-}$)	+1.8V to +11V
Junction Temperature (T_J)	-40°C to +85°C
Maximum Junction Temperature ($T_{J(MAX)}$) ⁽⁴⁾	+85°C
Package Thermal Resistance (θ_{JA}) ⁽⁵⁾	+252°C/W
Maximum Power Dissipation	Note 4

DC Electrical Characteristics

$V_{V+} = +1.8V$; $V_{V-} = 0V$; $V_{CM} = V_{OUT} = V_{V+}/2$; $R_L = 1M$; $T_J = +25^\circ C$, **bold** values indicate $-40^\circ C \leq T_J \leq +85^\circ C$, unless noted.

Symbol	Parameter	Condition	Min.	Typ.	Max.	Units
V_{OS}	Input Offset Voltage			0.9	7	mV
					9	
TCV_{OS}	Input Offset Voltage Temperature Drift			2.0		$\mu V/^\circ C$
I_B	Input Bias Current			1	10	pA
					500	
I_{OS}	Input Offset Current			0.01	0.5	pA
					75	
R_{IN}	Input Resistance			>10		T Ω
+PSRR	Positive Power Supply Rejection Ratio	$1.8V \leq V_{V+} \leq 5V, V_{V-} = 0V, V_{CM} = V_{OUT} = 0.9V$	60	85		dB
-PSRR	Negative Power Supply Rejection Ratio	$-1.8V \leq V_{V-} \leq -5V, V_{V+} = 0V, V_{CM} = V_{OUT} = -0.9V$	60	85		dB
CMRR	Common-Mode Rejection Ratio	$V_{CM} = -0.2V$ to $+2.0V$	50	70		dB
C_{IN}	Common-Mode Input Capacitance			3		pF
V_{OUT}	Output Voltage Swing	Output HIGH, $R_L = 100k$, Specified as $V_{V+} - V_{OUT}$		0.14	1	mV
					1	
		Output LOW, $R_L = 100k$		0.14	1	
					1	
		Output HIGH, $R_L = 2k$, Specified as $V_{V+} - V_{OUT}$		6.8	23	
					34	
		Output LOW, $R_L = 2k$		6.8	23	
					34	

Notes:

- Absolute maximum ratings indicate limits beyond which damage to the component may occur. Electrical specifications do not apply when operating the device outside its recommended operating ratings.
- The device is not guaranteed to function outside its operating ratings.
- I/O pin voltage is any external voltage to which an input or output is referenced.
- The maximum allowable power dissipation is a function of the maximum junction temperature, $T_{J(MAX)}$; the junction-to-ambient thermal resistance, θ_{JA} ; and the ambient temperature, T_A . The maximum allowable power dissipation at any ambient temperature is calculated using $P_D = (T_{J(MAX)} - T_A) \div \theta_{JA}$. Exceeding the maximum allowable power dissipation will result in excessive die temperature.
- Thermal resistance, θ_{JA} , applies to a part soldered on a printed-circuit board.
- Devices are ESD protected, however, handling precautions are recommended. All limits guaranteed by testing on statistical analysis. Human body model, 1.5k Ω in series with 100pF.

DC Electrical Characteristics (Continued)

$V_{V+} = +1.8V$; $V_{V-} = 0V$; $V_{CM} = V_{OUT} = V_{V+}/2$; $R_L = 1M$; $T_J = +25^{\circ}C$, **bold** values indicate $-40^{\circ}C \leq T_J \leq +85^{\circ}C$, unless noted.

Symbol	Parameter	Condition	Min.	Typ.	Max.	Units
I _{SC}	Output Short-Circuit Current ⁽⁷⁾	Sourcing, $V_{OUT} = 0V$	15	25		mA
		Sinking, $V_{OUT} = 1.8V$	15	25		
A _{VOL}	Voltage Gain	Sourcing		400		V/mV
		Sinking		400		
I _S	Supply Current	$V_{V+} = 1.8V$, $V_{OUT} = V_{V+}/2$		15	35	μA

AC Electrical Characteristics

$V_{V+} = +1.8V$; $V_{V-} = 0V$; $V_{CM} = V_{OUT} = V_{V+}/2$; $R_L = 1M$; $T_J = +25^{\circ}C$, **bold** values indicate $-40^{\circ}C \leq T_J \leq +85^{\circ}C$, unless noted.

Symbol	Parameter	Condition	Min.	Typ.	Max.	Units
SR	Slew Rate	Voltage follower, 1V step, $R_L = 100k$ @ 0.9V, $V_{OUT} = 1V_{P-P}$		0.015		V/ μs
GBW	Gain Bandwidth Product	Sourcing		25		kHz

DC Electrical Characteristics (2.7V)

$V_{V+} = +2.7V$; $V_{V-} = 0V$; $V_{CM} = V_{OUT} = V_{V+}/2$; $R_L = 1M$; $T_J = +25^{\circ}C$, **bold** values indicate $-40^{\circ}C \leq T_J \leq +85^{\circ}C$, unless noted.

Symbol	Parameter	Condition	Min.	Typ.	Max.	Units
V _{OS}	Input Offset Voltage			0.9	7	mV
				9		
TCV _{OS}	Input Offset Voltage Temperature Drift			2.0		$\mu V/^{\circ}C$
I _B	Input Bias Current			1	10	pA
				500		
I _{OS}	Input Offset Current			0.01	0.5	pA
				75		
R _{IN}	Input Resistance			>10		T Ω
+PSRR	Positive Power Supply Rejection Ratio	$2.7V \leq V_{V+} \leq 5V$, $V_{V-} = 0V$, $V_{CM} = V_{OUT} = 1.35V$	60	90		dB
-PSRR	Negative Power Supply Rejection Ratio	$-2.7V \leq V_{V-} \leq -5V$, $V_{V+} = 0V$, $V_{CM} = V_{OUT} = -1.35V$	60	90		dB
CMRR	Common-Mode Rejection Ratio	$V_{CM} = -0.2V$ to $+2.9V$	52	75		dB
C _{IN}	Common-Mode Input Capacitance			3		pF

Note:

7. Short circuit may cause the device to exceed maximum allowable power dissipation (see [Note 3](#)).

DC Electrical Characteristics (2.7V) (Continued)

$V_{V+} = +2.7V$; $V_{V-} = 0V$; $V_{CM} = V_{OUT} = V_{V+}/2$; $R_L = 1M$; $T_J = +25^{\circ}C$, **bold** values indicate $-40^{\circ}C \leq T_J \leq +85^{\circ}C$, unless noted.

Symbol	Parameter	Condition	Min.	Typ.	Max.	Units
V_{OUT}	Output Voltage Swing	Output HIGH, $R_L = 100k$, Specified as $V_{V+} - V_{OUT}$		0.2	1	mV
					1	
		Output LOW, $R_L = 100k$		0.2	1	
					1	
		Output HIGH, $R_L = 2k$, Specified as $V_{V+} - V_{OUT}$		10	33	
				50		
		Output LOW, $R_L = 2k$		10	33	
					50	
I_{SC}	Output Short-Circuit Current ⁽⁷⁾	Sourcing, $V_{OUT} = 0V$	30	50		mA
		Sinking, $V_{OUT} = 2.7V$	30	50		
A_{VOL}	Voltage Gain	Sourcing		400		V/mV
		Sinking		400		
I_S	Supply Current	$V_{V+} = 2.7V$, $V_{OUT} = V_{V+}/2$		17	42	μA

AC Electrical Characteristics (2.7V)

$V_{V+} = +2.7V$; $V_{V-} = 0V$; $V_{CM} = V_{OUT} = V_{V+}/2$; $R_L = 1M$; $T_J = +25^{\circ}C$, **bold** values indicate $-40^{\circ}C \leq T_J \leq +85^{\circ}C$, unless noted.

Symbol	Parameter	Condition	Min.	Typ.	Max.	Units
SR	Slew Rate	Voltage follower, 1V step, $R_L = 100k$ @ 1.35V, $V_{OUT} = 1V_{P-P}$		0.015		V/ μs
GBW	Gain Bandwidth Product	Sourcing		25		kHz

DC Electrical Characteristics (5V)

$V_{V+} = +5V$; $V_{V-} = 0V$; $V_{CM} = V_{OUT} = V_{V+}/2$; $R_L = 1M$; $T_J = +25^{\circ}C$, **bold** values indicate $-40^{\circ}C \leq T_J \leq +85^{\circ}C$, unless noted.

Symbol	Parameter	Condition	Min.	Typ.	Max.	Units
V_{OS}	Input Offset Voltage			0.9	7	mV
				9		
TCV_{OS}	Input Offset Voltage Temperature Drift			2.0		$\mu V/^{\circ}C$
I_B	Input Bias Current			1	10	pA
				500		
I_{OS}	Input Offset Current			0.01	0.5	pA
				75		
R_{IN}	Input Resistance			>10		T Ω
+PSRR	Positive Power Supply Rejection Ratio	$5V \leq V_{V+} \leq 10V$, $V_{V-} = 0V$, $V_{CM} = V_{OUT} = 2.5V$	65	95		dB
-PSRR	Negative Power Supply Rejection Ratio	$-5V \leq V_{V-} \leq -10V$, $V_{V+} = 0V$, $V_{CM} = V_{OUT} = -2.5V$	65	95		dB
CMRR	Common-Mode Rejection Ratio	$V_{CM} = -0.2V$ to $+5.2V$	57	80		dB
C_{IN}	Common-Mode Input Capacitance			3		pF
V_{OUT}	Output Voltage Swing	Output HIGH, $R_L = 100k$, Specified as $V_{V+} - V_{OUT}$		0.3	1.5	mV
				1.5		
		Output LOW, $R_L = 100k$		0.3	1.5	
				1.5		
		Output HIGH, $R_L = 2k$, Specified as $V_{V+} - V_{OUT}$		15	50	
				75		
		Output LOW, $R_L = 2k$		15	50	
				75		
I_{SC}	Output Short-Circuit Current ⁽⁷⁾	Sourcing, $V_{OUT} = 0V$	80	100		mA
		Sinking, $V_{OUT} = 5V$	80	100		
A_{VOL}	Voltage Gain	Sourcing		500		V/mV
		Sinking		500		
I_S	Supply Current	$V_{V+} = 5V$, $V_{OUT} = V_{V+}/2$		20	50	μA

AC Electrical Characteristics (5V)

$V_+ = +5V$; $V_- = 0V$; $V_{CM} = V_{OUT} = V_{V+}/2$; $R_L = 1M$; $T_J = +25^{\circ}C$, **bold** values indicate $-40^{\circ}C \leq T_J \leq +85^{\circ}C$, unless noted.

Symbol	Parameter	Condition	Min.	Typ.	Max.	Units
SR	Slew Rate	Voltage follower, 1V step, $R_L = 100k$ @ 1.5V, $V_{OUT} = 1V_{P-P}$		0.02		V/ μs
GBW	Gain Bandwidth Product	Sourcing		25		kHz

DC Electrical Characteristics (10V)

$V_{V+} = +10V$; $V_{V-} = 0V$; $V_{CM} = V_{OUT} = V_{V+}/2$; $R_L = 1M$; $T_J = +25^{\circ}C$, **bold** values indicate $-40^{\circ}C \leq T_J \leq +85^{\circ}C$, unless noted.

Symbol	Parameter	Condition	Min.	Typ.	Max.	Units
V_{OS}	Input Offset Voltage			0.9	7	mV
					9	
TCV_{OS}	Input Offset Voltage Temperature Drift			2.0		$\mu V/^{\circ}C$
I_B	Input Bias Current			1	10	pA
					500	
I_{OS}	Input Offset Current			0.01	0.5	pA
					75	
R_{IN}	Input Resistance			>10		T Ω
+PSRR	Positive Power Supply Rejection Ratio	$5V \leq V_{V+} \leq 10V$, $V_{V-} = 0V$, $V_{CM} = V_{OUT} = 2.5V$	65	95		dB
-PSRR	Negative Power Supply Rejection Ratio	$-5V \leq V_{V-} \leq -10V$, $V_{V+} = 0V$, $V_{CM} = V_{OUT} = -2.5V$	65	95		dB
CMRR	Common-Mode Rejection Ratio	$V_{CM} = -0.2V$ to $+10.2V$	60	85		dB
C_{IN}	Common-Mode Input Capacitance			3		pF
V_{OUT}	Output Voltage Swing	Output HIGH, $R_L = 100k$, Specified as $V_{V+} - V_{OUT}$		0.45	2.5	mV
					2.5	
		Output LOW, $R_L = 100k$		0.45	2.5	
					2.5	
		Output HIGH, $R_L = 2k$, Specified as $V_{V+} - V_{OUT}$		24	80	
					120	
		Output LOW, $R_L = 2k$		24	80	
					120	
I_{SC}	Output Short-Circuit Current ⁽⁷⁾	Sourcing, $V_{OUT} = 0V$	100	200		mA
		Sinking, $V_{OUT} = 10V$	100	200		
A_{VOL}	Voltage Gain	Sourcing		500		V/mV
		Sinking		500		
I_S	Supply Current	$V_{V+} = 10V$, $V_{OUT} = V_{V+}/2$		25	65	μA

AC Electrical Characteristics (10V)

$V_+ = +10V$; $V_- = 0V$; $V_{CM} = V_{OUT} = V_{V+}/2$; $R_L = 1M$; $T_J = +25^\circ C$, **bold** values indicate $-40^\circ C \leq T_J \leq +85^\circ C$, unless noted.

Symbol	Parameter	Condition	Min.	Typ.	Max.	Units
SR	Slew Rate	Voltage follower, 1V step, $R_L = 100k$ @ 1.35V $V_{OUT} = 1V_{P-P}$		0.02		V/ μs
GBW	Gain Bandwidth Product			25		kHz
ϕ_M	Phase Margin			50		$^\circ$
G_M	Gain Margin			15		dB
e_N	Input-Referred Voltage Noise	$f = 1kHz$, $V_{CM} = 1.0V$		110		nV/\sqrt{Hz}
i_N	Input-Referred Current Noise	$f = 1kHz$		0.03		pA/\sqrt{Hz}

Application Information

Input Common Mode Voltage

The MIC7111 tolerates input overdrive by at least 300mV beyond either rail without producing phase inversion.

If the absolute maximum input voltage is exceeded, the input current should be limited to $\pm 5\text{mA}$ maximum to prevent reducing reliability. A $10\text{k}\Omega$ series input resistor, used as a current limiter, will protect the input structure from voltages as large as 50V above the supply or below ground. See [Figure 1](#).

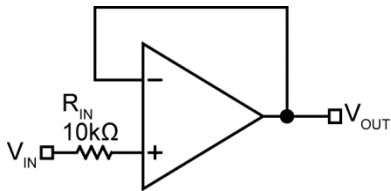


Figure 1. Input Current-Limit Protection

Output Voltage Swing

Sink and source output resistances of the MIC7111 are equal. Maximum output voltage swing is determined by the load and the approximate output resistance. The output resistance is presented in Equation 1:

$$R_{OUT} = \frac{V_{DROP}}{I_{LOAD}} \quad \text{Eq. 1}$$

V_{DROP} is the voltage dropped within the amplifier output stage. V_{DROP} and I_{LOAD} can be determined from the V_O (output swing) portion of the appropriate electrical characteristics table. I_{LOAD} is equal to the typical output high voltage minus $V+/2$ and divided by R_{LOAD} . For example, using the *DC Electrical Characteristics (5V)* table, the typical output voltage drop using a $2\text{k}\Omega$ load (connected to $V+/2$) is 0.015V , which produces an I_{LOAD} of:

$$\frac{2.5\text{V} - 0.015\text{V}}{2\text{k}\Omega} = 1.243\text{mA} \quad \text{Eq. 2}$$

Then,

$$R_{OUT} = \frac{15\text{mV}}{1.243\text{mA}} = 12.1 = 12\Omega \quad \text{Eq. 3}$$

Driving Capacitive Loads

Driving a capacitive load introduces phase-lag into the output signal, and this in turn reduces op-amp system phase margin. The application that is least forgiving of reduced phase margin is a unity gain amplifier. The MIC7111 can typically drive a 500pF capacitive load connected directly to the output when configured as a unity-gain amplifier.

Using Large-Value Feedback Resistors

A large-value feedback resistor ($> 500\text{k}\Omega$) can reduce the phase margin of a system. This occurs when the feedback resistor acts in conjunction with input capacitance to create phase lag in the feedback signal. Input capacitance is usually a combination of input circuit components and other parasitic capacitance, such as amplifier input capacitance and stray printed circuit board capacitance.

[Figure 2](#) illustrates a method of compensating phase lag caused by using a large-value feedback resistor. Feedback capacitor C_{FB} introduces sufficient phase lead to overcome the phase lag caused by feedback resistor R_{FB} and input capacitance C_{IN} . The value of C_{FB} is determined by first estimating C_{IN} and then applying the following formula:

$$R_{IN} \times C_{IN} \leq R_{FB} \times C_{FB} \quad \text{Eq. 4}$$

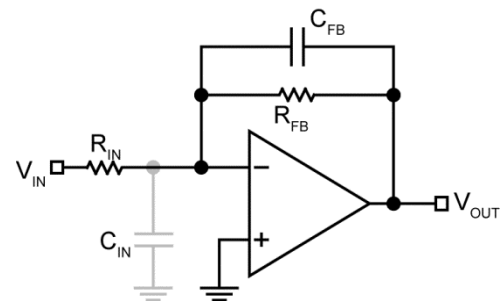


Figure 2. Cancelling Feedback Phase Lag

Since a significant percentage of C_{IN} may be caused by board layout, it is important to note that the correct value of C_{FB} may change when changing from a breadboard to the final circuit layout.

Typical Circuits

Some single-supply, rail-to-rail applications – for which the MIC7111 is well suited – are shown in the circuit diagrams of Figures 3 through 8.

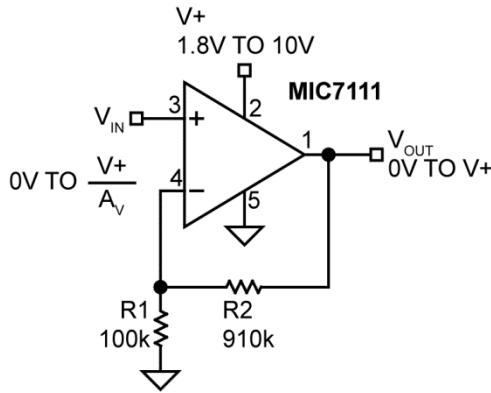


Figure 3. Noninverting Amplifier

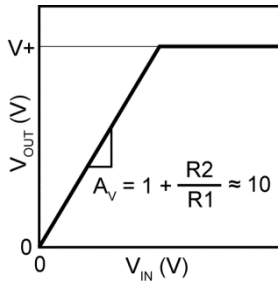


Figure 4. Noninverting Amplifier Behavior

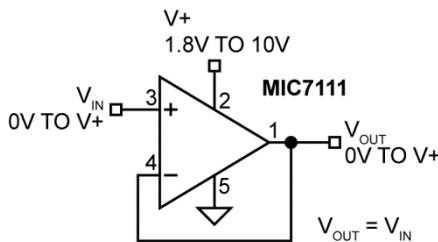


Figure 5. Voltage Follower/Buffer

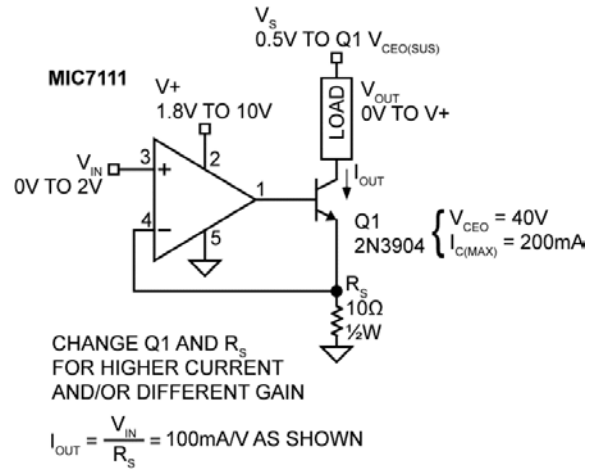


Figure 6. Voltage-Controlled Current Sink

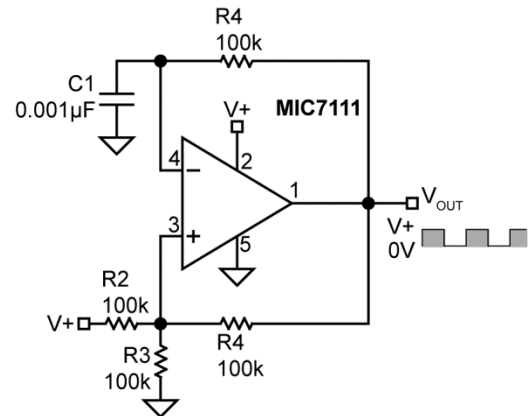


Figure 7. Square Wave Oscillator

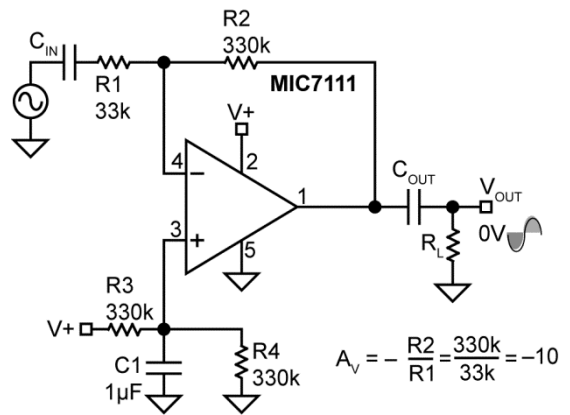
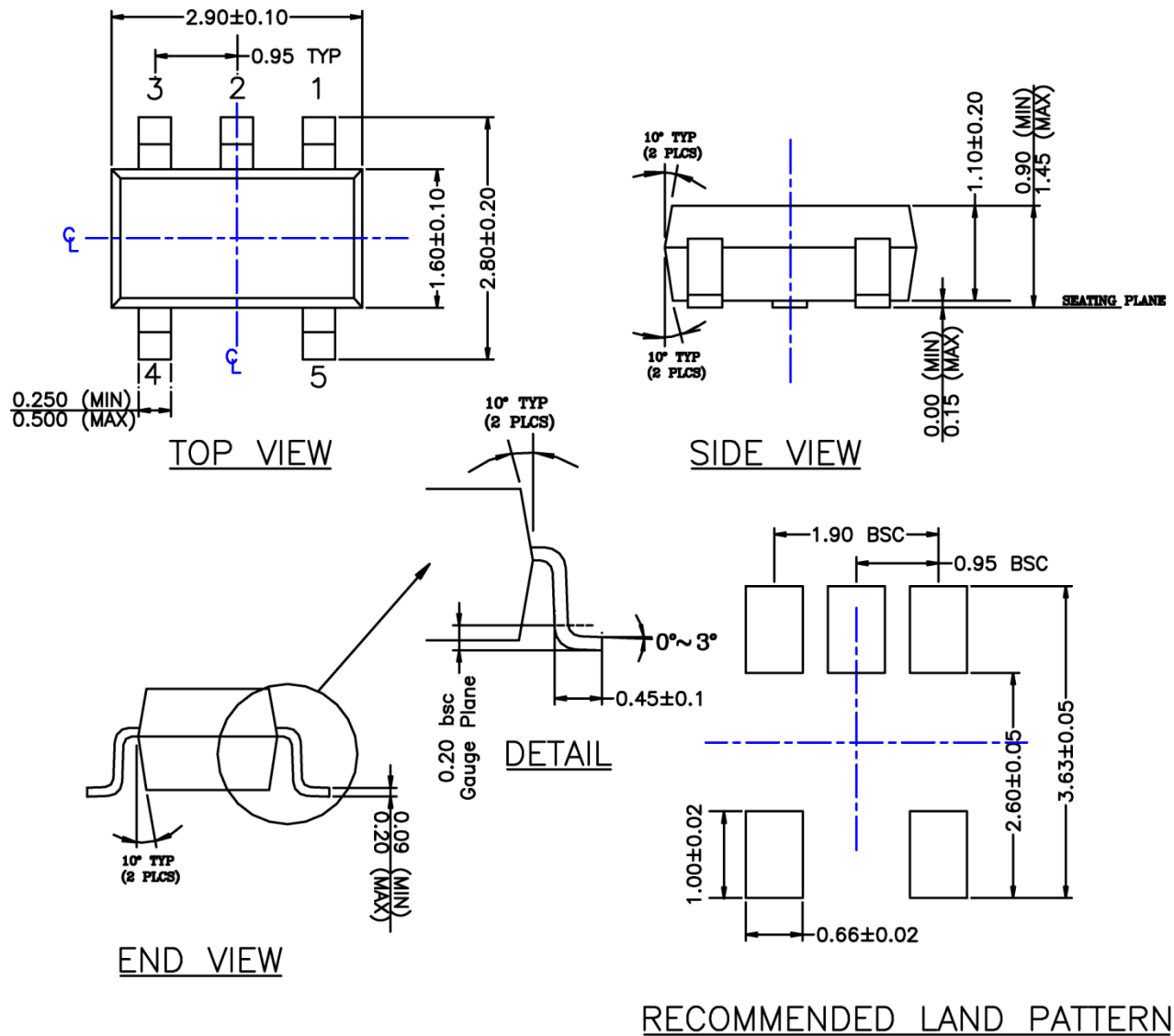


Figure 8. AC-Coupled Inverting Amplifier

Package Information⁽¹⁾ and Recommended Landing Pattern



- NOTE:**
1. PACKAGE OUTLINE EXCLUSIVE OF MOLD FLASH & BURR.
 2. PACKAGE OUTLINE INCLUSIVE OF SOLER PLATING.
 3. DIMENSION AND TOLERANCE PER ANSI Y14.5M, 1982.
 4. FOOT LENGTH MEASUREMENT BASED ON GAUGE PLANE METHOD.
 5. DIE FACES UP FOR MOLD, AND FACES DOWN FOR TRIM/FORM.
 6. ALL DIMENSIONS ARE IN MILLIMETERS.

SOT23-5 (M5)

Note:

1. Package information is correct as of the publication date. For updates and most current information, go to www.micrel.com.

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