**Features** 



## Quad PCI Express, Hot-Plug Controllers

### **General Description**

The MAX5959/MAX5960 quad hot-plug controllers are designed for PCI Express<sup>®</sup> (PCIe) applications. These devices provide hot-plug control for the main 12V, 3.3V, and 3.3V auxiliary supplies of four PCIe slots. The MAX5959/MAX5960s' logic inputs/outputs allow interfacing directly with the system hot-plug management controller or through an SMBus<sup>™</sup> with an external I/O expander such as the MAX7313. An integrated debounced attention switch and present-detect signals simplify system design.

The MAX5959/MAX5960 drive eight external n-channel MOSFETs to control the 12V and 3.3V main outputs. The 3.3V auxiliary outputs are controlled through  $0.2\Omega$  n-channel MOSFETs. Internal charge pumps provide the gate drive for the 12V outputs while the gate drive of the 3.3V output is driven by the 12V input supply clamped to 5.5V above the respective 3.3V main supply rail. The 3.3V auxiliary outputs are completely independent from the main outputs with their own charge pumps.

At power-up, the MAX5959/MAX5960 keep all the MOSFETs off until the supplies rise above their respective undervoltage lockout (UVLO) thresholds. Upon a turn-on command, the MAX5959/MAX5960 enhance the external and internal MOSFETs slowly with a constant gate current to limit the power-supply inrush current.

The MAX5959/MAX5960 actively limit the current to protect all outputs at all times and shut down if an overcurrent condition persists longer than the programmable overcurrent timeout. After an overcurrent or overtemperature fault condition, the MAX5959L/MAX5960L latch off while the MAX5959A/MAX5960A automatically restart after a restart time delay. The MAX5959/MAX5960 are offered in latch-off or autorestart versions (see the *Selector Guide*).

The MAX5959/MAX5960 are available in an 80-pin TQFP package and operate over the -40°C to +85°C temperature range.

\_Applications

Servers

Desktop Mobile Server Platforms

Workstations

**♦ PCle Compliant** 

- ♦ Hot Swap 12V, 3.3V, and 3.3V Auxiliary for 4 PCle Slots
- ♦ Integrated Power MOSFETs for Auxiliary Supply Rails
- ♦ Controls di/dt and dV/dt
- ♦ Active Current Limiting Protects Against Overcurrent/Short-Circuit Conditions
- ♦ Programmable Current-Limit Timeout
- PWRGD Signal Outputs with Programmable Power-On Reset (POR) (160ms Default)
- ◆ Latched FAULT Signal Output After Overcurrent or Overtemperature Fault
- Attention Switch Inputs/Outputs with 4ms Debounce
- **♦ Present-Detect Inputs**
- ♦ Force-On Inputs Facilitate Testing/Debug
- **♦ Thermal Shutdown**
- ♦ Allow Control Through SMBus with an I/O Expander

### **Ordering Information**

PART	TEMP RANGE	PIN- PACKAGE	PKG CODE
MAX5959AECS+	-40°C to +85°C	80 TQFP	C80-1
MAX5959LECS+	-40°C to +85°C	80 TQFP	C80-1
MAX5960AECS+	-40°C to +85°C	80 TQFP	C80-1
MAX5960LECS+	-40°C to +85°C	80 TQFP	C80-1

<sup>+</sup>Denotes a lead-free package.

Pin Configuration and Typical Application Circuit appear at end of data sheet.

PCI Express is a registered trademark of PCI-SIG Corp. SMBus is a trademark of Intel Corp.

### **Selector Guide**

PART	LATCH OFF	AUTORESTART	GUARANTEED AUX CURRENT (mA)
MAX5959AECS+		$\checkmark$	375
MAX5959LECS+	V		375
MAX5960AECS+		V	550
MAX5960LECS+	V		550

Maxim Integrated Products

### ABSOLUTE MAXIMUM RATINGS

(All voltages referenced to GND,	unless otherwise noted.)
12VIN	0.3V to +14V
12G	0.3V to $(V_{12VIN} + 6V)$
12S_+, 12S, 3.3G	0.3V to $(V_{12VIN} + 0.3V)$
3.3VAUXIN, ON_, FAULT_, PWRO	GD0.3V to +6V
PGND	0.3V to +0.3V
All Other Pins	0.3V to $(V_{3.3VAUXIN} + 0.3V)$

Continuous Power Dissipation ( $T_A = +70^{\circ}C$ )	
80-Pin TQFP (derate 23.3mW/°C above +70	°C)1860mW
Operating Temperature Range	40°C to +85°C
Junction Temperature	+150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### **ELECTRICAL CHARACTERISTICS**

 $\begin{array}{l} (V_{12VIN}=V_{12S\_+}=V_{12S\_-}=12V,\ V_{3.3S\_+}=V_{3.3S\_-}=V_{3.3\underline{AUXIN}}=\underline{V_{ON}}=V_{AUXON}=V_{\overline{FON}}=3.3V,\ \overline{PWRGD}=\overline{FAULT}=PORADJ=12G\_=12G\_=3.3G\_=OPEN,\ INPUT\_=PRES-DET\_=PGND=GND,\ T_A=T_J=-40^{\circ}C\ to\ +85^{\circ}C,\ unless\ otherwise\ noted.\ Typical\ values\ are\ at\ T_A=T_J=+25^{\circ}C.)\ (Note\ 1) \end{array}$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
12V SUPPLY (MAIN)							
12V Supply Input-Voltage Range	V <sub>12VIN</sub>		10.8	12	13.2	V	
12V/ Indervolte de Legiscut	V	V <sub>12VIN</sub> rising	9.5	10	10.5	V	
12V Undervoltage Lockout	V <sub>12UVLO</sub>	Hysteresis		0.1		1 V	
12V Undervoltage Lockout Deglitch Time	tDEG,UVLO	V <sub>12IN</sub> falling below V <sub>12UVLO</sub> to UVLO assertion		30		μs	
12V Supply Current	I <sub>12VIN</sub>	V <sub>12VIN</sub> = 13.2V		1	2.5	mA	
12VIN CONTROL							
12VIN Current-Limit Threshold (V <sub>12S+</sub> - V <sub>12S-</sub> )	V <sub>12ILIM</sub>		49	54	59	mV	
12G_ Gate Charge Current	I <sub>12</sub> G, CHG	V <sub>12G</sub> _ = GND	4	5	6	μΑ	
		Normal turn-off, $ON_{-} = GND$ , $V_{12G_{-}} = 2V$	50	150	250	μΑ	
12G_ Gate Discharge Current	I <sub>12G_</sub> , DIS	Output short-circuit condition, strong gate pulldown to regulation, $(V_{12S+} - V_{12S-}) \ge 1V$ , $V_{12G-} = 5V$	60	120	200	mA	
12G_ Gate High Voltage (V12G V12VIN)	V <sub>12G, H</sub>	I <sub>12G</sub> _ = 1μA	4.8	5.3	5.8	V	
12G_ Threshold Voltage for PWRGD_ Assertion	VPGTH12	Referred to V <sub>12VIN</sub> , I <sub>12G</sub> = 1µA (Note 2)	3	4	4.8	V	
12S Input Bias Current					1	μΑ	
12S_+ Input Bias Current				10	30	μΑ	
3.3V SUPPLY (MAIN)							
3.3V Supply Voltage Range	V <sub>3.3S_+</sub>		3.0	3.3	3.6	V	
Undervoltage Lockout (Note 3)		3.3SA+ rising	2.50	2.65	2.78	V	
Undervollage Lockout (Note 3)		Hysteresis		30		mV	

### **ELECTRICAL CHARACTERISTICS (continued)**

 $(V_{12VIN} = V_{12S\_+} = V_{12S\_-} = 12V, V_{3.3S\_+} = V_{3.3S\_-} = V_{3.3AUXIN} = V_{ON\_} = V_{AUXON\_} = V_{FON\_} = 3.3V, \overline{PWRGD\_} = \overline{FAULT\_} = PORADJ = TIM = OUTPUT\_ = 12G\_ = 3.3G\_ = OPEN, INPUT\_ = PRES-DET\_ = PGND = GND, TA = TJ = -40°C to +85°C, unless otherwise noted. Typical values are at TA = TJ = +25°C.) (Note 1)$ 

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
3.3V CONTROL							
3.3V Current-Limit Threshold (V <sub>3.3S_+</sub> - V <sub>3.3S</sub> )	V3.3ILIM			17	20	23	mV
3.3G_ Gate Charge Current	I <sub>3.3G_,CHG</sub>	V <sub>3.3G</sub> _ = GND		4	5	6	μΑ
		Normal turn-off, ON_ = GI V <sub>3.3G</sub> _ = 2V	ND,	50	150	250	μΑ
3.3G_ Gate Discharge Current	l3.3G_, DIS	Output short-circuit condi gate pulldown to regulatio V <sub>3.3S</sub> _+ - V <sub>3.3S</sub> ≥ 1V, V <sub>3</sub>	on,	90	150	250	mA
3.3G_ Gate High Voltage (V3.3G V3.3S_+)	V3.3G_H	ISOURCE = 1µA		4.5	5.5	6.8	V
3.3G_ Threshold Voltage for PWRGD_ Assertion	VPGTH3.3	Referred to V <sub>3.3AUXIN</sub> , I <sub>3.3</sub> (Note 2)	3G_ = 1µA	3	4	4.5	V
3.3S Input Bias Current						1	μΑ
3.3S_+ Input Bias Current					20	60	μΑ
3.3V AUXILIARY SUPPLY							
3.3VAUXIN Supply Input Voltage Range	V <sub>3.3AUXIN</sub>			3.0	3.3	3.6	V
3.3VAUXIN Undervoltage	M	V <sub>3.3VAUXIN</sub> rising Hysteresis		2.50	2.65	2.78	V
Lockout	V3.3VAUXUVLO				30		mV
3.3VAUXIN Supply Current		V3.3VAUXIN = 3.6V			2.5	5	mA
3.3VAUXIN to 3.3VAUXO_		I <sub>3.3VAUXO</sub> _ = 550mA (MA	X5960)			280	mV
Maximum Dropout		I <sub>3.3VAUXO</sub> _ = 375mA (MA	X5959)			225	IIIV
3.3VAUXO_ Current-Limit		3.3VAUXO_ shorted to	MAX5959	376	450	564	mA
Threshold		GND	MAX5960	560	700	850	ША
3.3VAUXO_ Threshold for PWRGD_ Assertion (V3.3VAUXIN - V3.3VAUXO_)	VPGTH3.3AUX	(Note 2)				400	mV
LOGIC SIGNALS							
Input-Logic Threshold (ON_, FON_, AUXON_, PRES-DET_,		Rising edge		1.0		2.0	V
INPUT_)		Hysteresis			25		mV
Input Bias Current (ON_, AUXON_, INPUT_)						1	μΑ
FON_, PRES-DET_ Internal				25	50	75	kΩ
ON_, AUXON_ High-to-Low Deglitch Time					4		μs

### **ELECTRICAL CHARACTERISTICS (continued)**

 $(V_{12VIN} = V_{12S\_+} = V_{12S\_-} = 12V, V_{3.3S\_+} = V_{3.3S\_-} = V_{3.3AUXIN} = V_{ON\_} = V_{AUXON\_} = V_{FON\_} = 3.3V, \overline{PWRGD\_} = \overline{FAULT\_} = PORADJ = TIM = OUTPUT\_ = 12G\_ = 3.3G\_ = OPEN, INPUT\_ = PRES-DET\_ = PGND = GND, TA = TJ = -40°C to +85°C, unless otherwise noted. Typical values are at TA = TJ = +25°C.) (Note 1)$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
PRES-DET_ High-to-Low Deglitch Time	tDEG		3	5	7	ms
		V <sub>PORADJ</sub> = GND			1	μs
DWDOD D		PORADJ = unconnected	90	160	250	
PWRGD_ Power-On Reset Deglitch Time (Note 2)	tpor_hl	$R_{PORADJ} = 20k\Omega$	35	55	75	ms
Bogillon Time (Note 2)		$R_{PORADJ} = 100k\Omega$	145	265	380	1115
		$R_{PORADJ} = 200k\Omega$		530		
PWRGD_ Low-to-High Deglitch Time	tpor_lh			4		μs
PWRGD_, FAULT_ Output Low		I <sub>SINK</sub> = 2mA			0.1	V
Voltage		I <sub>SINK</sub> = 30mA			0.8	V
PWRGD_, FAULT_ Output High Leakage Current		PWRGD_, FAULT_ = 5.5V			1	μΑ
		TIM = open	5.5	11	17	
FAULT_ Timeout	<b>*</b>	$R_{TIM} = 15k\Omega$	1.4	2.6	3.8	ms
FAULT_TIMEOUT	tfault	$R_{TIM} = 120k\Omega$	12	22	32	
		$R_{TIM} = 300k\Omega$		53		
FAULT_ Timeout During Startup	tsu			2 x t <sub>FAULT</sub>		ms
Autorestart Delay Time	<sup>†</sup> RESTART			64 x t <sub>FAULT</sub>		ms
Fault Reset Minimum Pulse Width	treset	(Note 4)		100		μs
Thermal Shutdown Threshold	tsD	T <sub>J</sub> rising		150		°C
Thermal Shutdown Hysteresis				20		°C
OUT_ Debounce Time	t <sub>DBC</sub>		2.6	4.4	6.2	ms
OUT_ Voltage High		ISOURCE = 2mA	V3.3AUXIN - 0.3	1	V <sub>3.3AUXIN</sub>	V

Note 1: All devices are 100% production tested at  $T_A = +25$ °C. Limits over temperature are guaranteed by design.

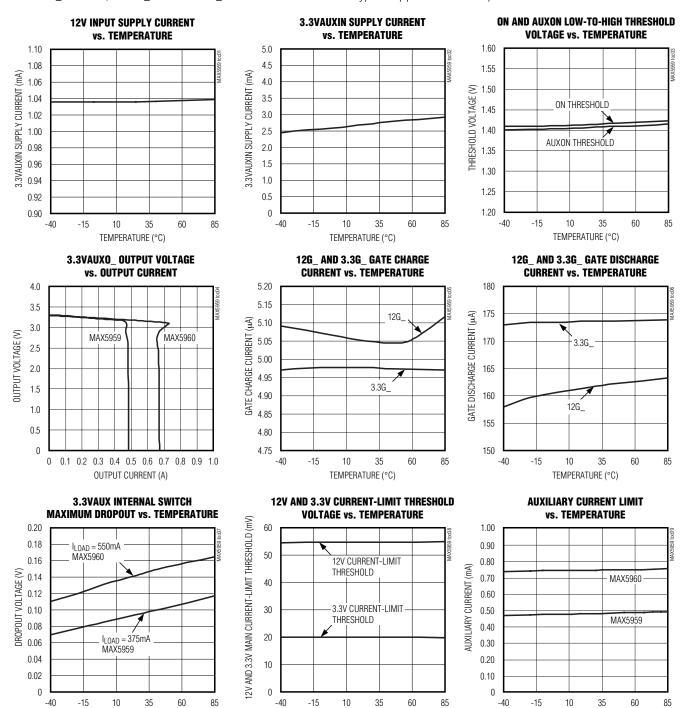
Note 2: PWRGD\_ asserts a time tpor\_HL after VpgTH12, VpgTH3.3, and VpgTH3.3AUX conditions are met.

Note 3: The UVLO for the 3.3V supply is sensed at 3.3SA+.

Note 4: This is the time that ON\_ or AUXON\_ must stay low when resetting a fault condition.

### **Typical Operating Characteristics**

 $(V_{12VIN} = V_{12S_+} = 12V, V_{3.3S_+} = V_{3.3S_-} = V_{3.3AUXIN} = V_{ON_-} = V_{AUXON_-} = V_{\overline{FON_-}} = 3.3V, \overline{PWRGD_-} = \overline{FAULT_-} = PORADJ = TIM = OUTPUT_- = OPEN, INPUT_- = PRES-DET_- = PGND = GND. See the$ *Typical Application Circuit.*)



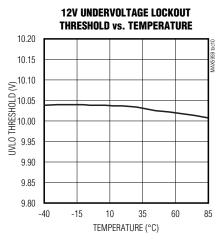
TEMPERATURE (°C)

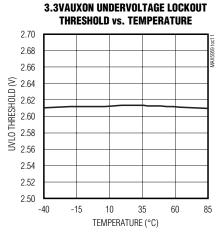
TEMPERATURE (°C)

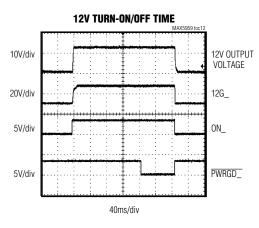
TEMPERATURE (°C)

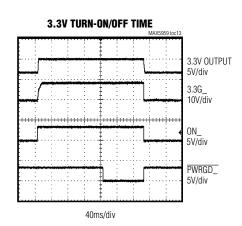
### Typical Operating Characteristics (continued)

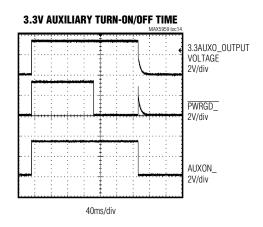
 $(V_{12VIN} = V_{12S\_+} = 12V, V_{3.3S\_+} = V_{3.3S\_-} = V_{3.3AUXIN} = V_{ON\_} = V_{AUXON\_} = V_{\overline{FON\_}} = 3.3V, \overline{PWRGD\_} = \overline{FAULT\_} = PORADJ = TIM = OUTPUT\_ = OPEN, INPUT\_ = PRES-DET\_ = PGND = GND. See the$ *Typical Application Circuit.*)

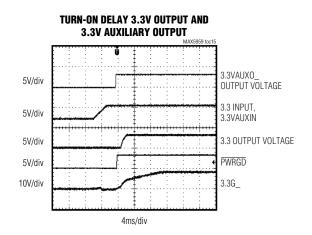


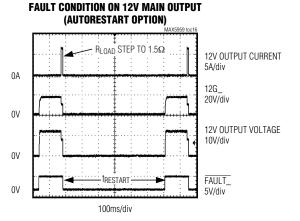








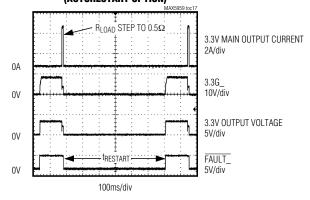




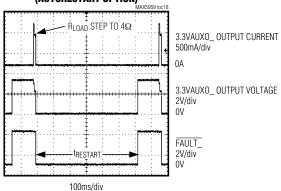
### Typical Operating Characteristics (continued)

 $(V_{12VIN} = V_{12S_+} = 12V, V_{3.3S_+} = V_{3.3S_-} = V_{3.3AUXIN} = V_{ON_-} = V_{AUXON_-} = V_{\overline{FON_-}} = 3.3V, \overline{PWRGD_-} = \overline{FAULT_-} = PORADJ = TIM = OUTPUT_- = OPEN, INPUT__ = PRES-DET__ = PGND = GND. See the$ *Typical Application Circuit*.)

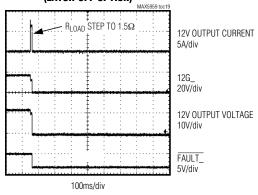
## FAULT CONDITION ON 3.3V MAIN OUTPUT (AUTORESTART OPTION)



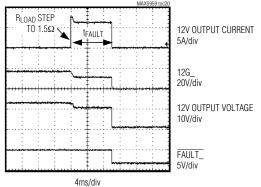
#### FAULT CONDITION ON AUXILIARY OUTPUT (AUTORESTART OPTION)



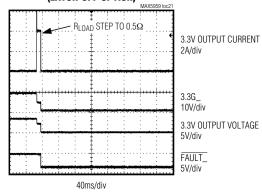
## FAULT CONDITION ON 12V MAIN OUTPUT (LATCH-OFF OPTION)



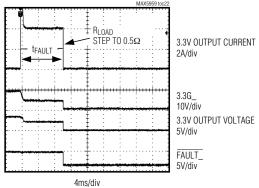
### **FAULT CONDITION ON 12V OUTPUT**



## FAULT CONDITION ON 3.3V MAIN OUTPUT (LATCH-OFF OPTION)



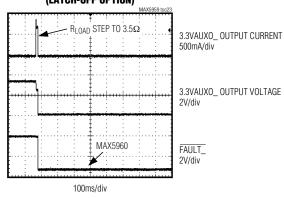
### **FAULT CONDITION ON 3.3V MAIN OUTPUT**



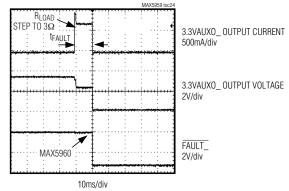
### Typical Operating Characteristics (continued)

 $(V_{12VIN} = V_{12S_+} = 12V, V_{3.3S_+} = V_{3.3S_-} = V_{3.3AUXIN} = V_{ON_-} = V_{AUXON_-} = V_{\overline{FON_-}} = 3.3V, \overline{PWRGD_-} = \overline{FAULT_-} = PORADJ = TIM = OUTPUT_- = OPEN, INPUT__ = PRES-DET__ = PGND = GND. See the$ *Typical Application Circuit*.)

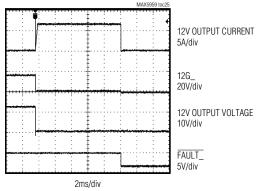
## FAULT CONDITION ON AUXILIARY OUTPUT (LATCH-OFF OPTION)



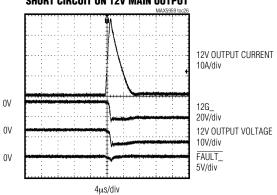
### **FAULT CONDITION ON AUXILIARY OUTPUT**



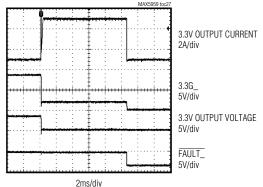
#### **SHORT CIRCUIT ON 12V MAIN OUTPUT**



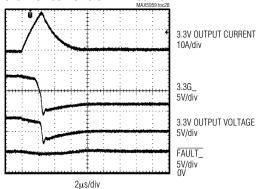
### **SHORT CIRCUIT ON 12V MAIN OUTPUT**



#### **SHORT CIRCUIT ON 3.3V MAIN OUTPUT**



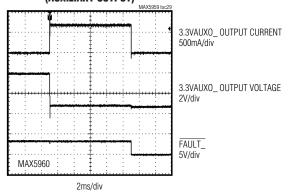
#### **SHORT CIRCUIT ON 3.3V MAIN OUTPUT**



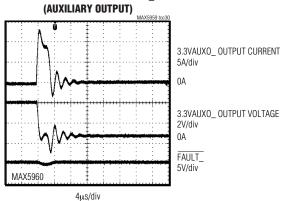
### Typical Operating Characteristics (continued)

 $(V_{12VIN}=V_{12S\_+}=12V,\ V_{3.3S\_+}=V_{3.3S\_-}=V_{3.3AUXIN}=V_{ON\_}=V_{AUXON\_}=V_{\overline{FON\_}}=3.3V,\ \overline{PWRGD}\_=\overline{FAULT}\_=PORADJ=TIM=V_{AUXON\_}=V_{\overline{FON\_}}=3.3V$ OUTPUT = OPEN, INPUT = PRES-DET = PGND = GND. See the Typical Application Circuit.)

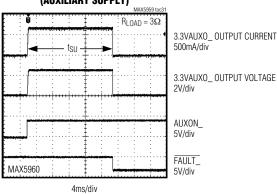
### SHORT CIRCUIT ON 3.3AUXO\_ (AUXILIARY OUTPUT)



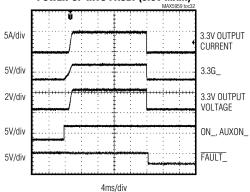
## SHORT CIRCUIT ON 3.3VAUXO\_



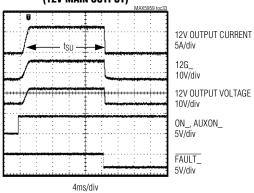
#### **POWER-UP INTO FAULT** (AUXILIARY SUPPLY)



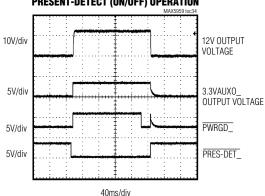
#### **POWER-UP INTO FAULT (3.3V MAIN)**



#### **POWER-UP INTO FAULT** (12V MAIN OUTPUT)

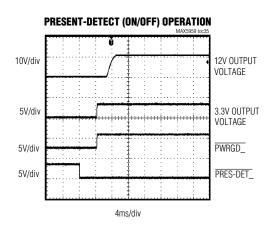


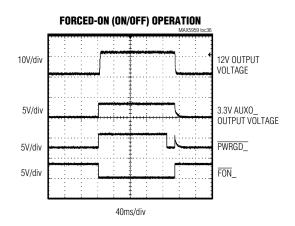
#### PRESENT-DETECT (ON/OFF) OPERATION

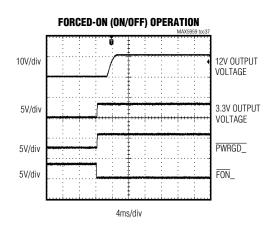


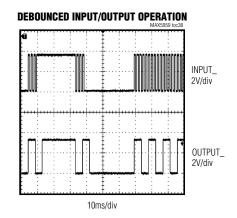
### Typical Operating Characteristics (continued)

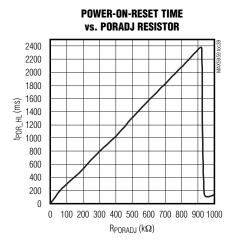
 $(V_{12VIN} = V_{12S_+} = 12V, V_{3.3S_+} = V_{3.3S_-} = V_{3.3AUXIN} = V_{ON_-} = V_{AUXON_-} = V_{\overline{FON_-}} = 3.3V, \overline{PWRGD_-} = \overline{FAULT_-} = PORADJ = TIM = OUTPUT_- = OPEN, INPUT_- = PRES-DET_- = PGND = GND. See the$ *Typical Application Circuit.*)

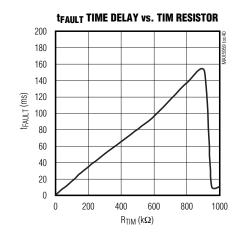












## **Pin Description**

		Pili Description
PIN	NAME	FUNCTION
1	12SB-	Slot B 12V Negative Current-Sense Input. Connect 12SB- to the negative side of the current-sense resistor using the Kelvin-sensing technique to ensure accurate current sensing.
2	12SB+	Slot B 12V Positive Current-Sense Input. Connect the positive terminal of the current-sense resistor to 12SB+ using the Kelvin-sensing technique to ensure accurate current sensing.
3	AUXONB	Slot B 3.3V Auxiliary Output Enable. A logic-high at AUXONB turns on the slot B auxiliary output.
4	ONB	Slot B 12V and 3.3V Main Outputs Enable. A logic-high at ONB turns on the 12V and 3.3V main outputs of slot B (see Table 2).
5	INPUT1	Digital Logic Gate Input 1
6	OUTPUT1	Digital Output 1. 4ms debounced digital output of INPUT1.
7	INPUT2	Digital Logic Gate Input 2
8	OUTPUT2	Digital Output 2. 4ms debounced digital output of INPUT2.
9, 61, 80	N.C.	No Connection. Not internally connected. Leave unconnected.
10	FOND	Slot D Forced-On Input. $\overline{\text{FOND}}$ has a 50k $\Omega$ internal pullup to 3.3VAUXIN. A logic-low on $\overline{\text{FOND}}$ turns on all slot D outputs as long as the power inputs are within their operating range, regardless of the status of the other input signals. Leave $\overline{\text{FOND}}$ unconnected for normal operation.
11	PRES-DETD	Slot D Present-Detect Input. $\overline{\text{PRES-DETD}}$ accepts inputs from PRSNT#_# on a PCIe connector. $\overline{\text{PRES-DETD}}$ has an internal $50\text{k}\Omega$ pullup to $3.3\text{VAUXIN}$ . When $\overline{\text{PRES-DETA}}$ is low, the outputs follow the command from OND and AUXOND after a 4ms debounced time. When $\overline{\text{PRES-DETD}}$ goes from low to high, all outputs of the respective slot shut down with no delay.
12	OND	Slot D 12V and 3.3V Main Outputs Enable. A logic-high at OND turns on the 12V and 3.3V main outputs of slot D (see Table 2).
13	AUXOND	Slot D 3.3V Auxiliary Output Enable. A logic-high at AUXOND turns on the slot D auxiliary output.
14	GND	Ground
15	FONC	Slot C Forced-On Input C. FONC has a 50kΩ internal pullup to 3.3VAUXIN. A logic-low on FONC turns on all slot C outputs as long as the power inputs are within their operating range, regardless of the status of the other input signals. Leave FONC unconnected for normal operation.
16	PRES-DETC	Slot C Present-Detect Input. $\overline{\text{PRES-DETC}}$ accepts inputs from PRSNT#_# on a PCIe connector. $\overline{\text{PRES-DETC}}$ has an internal $50\text{k}\Omega$ pullup to 3.3VAUXIN. When $\overline{\text{PRES-DETC}}$ is low, the outputs follow the command from ONC and AUXONC after a 4ms debounced time. When $\overline{\text{PRES-DETC}}$ goes from low to high, all outputs of the respective slot shut down with no delay.
17	ONC	Slot C 12V And 3.3V Main Outputs Enable. A logic-high at ONC turns on the 12V and 3.3V main outputs of slot C (see Table 2).
18	AUXONC	Slot C 3.3V Auxiliary Output Enable. A logic-high at AUXONC turns on the slot C auxiliary output.
19	12SC+	Slot C 12V Positive Current-Sense Input. Connect the positive terminal of the current-sense resistor to 12SC+ using the Kelvin-sensing technique to ensure accurate current sensing.
20	12SC-	Slot C 12V Negative Current-Sense Input. Connect 12SC- to the negative side of the current-sense resistor using the Kelvin-sensing technique to ensure accurate current sensing.
21	12GC	Slot C 12V Gate-Drive Output. Connect 12GC to the gate of slot C's 12V MOSFET. At power-up, 12GC is raised to the internal charge-pump voltage level by a constant current.
22	3.3SC+	Slot C 3.3V Positive Current-Sense Input. Connect the positive side of the current-sense resistor to 3.3SC+ using the Kelvin-sensing technique to ensure accurate current sensing.

## \_Pin Description (continued)

PIN	NAME	FUNCTION
23	3.3SC-	Slot C 3.3V Negative Current-Sense Input. Connect to the negative side of the sense resistor using the Kelvin-sensing technique to ensure accurate current sensing.
24	3.3GC	Slot C 3.3V Gate-Drive Output. Connect 3.3GC to the gate of slot C's 3.3V MOSFET. At power-up, 3.3GC is charged to 5.5V above the 3.3V supply by a constant current derived from V <sub>12VIN</sub> . V <sub>3.3GC</sub> 's rise time is determined by the external gate capacitance.
25	FAULTC	Open-Drain Fault Output Signal. FAULTC latches active-low whenever slot C outputs are shut down due to a fault. A fault is either of:  • An overcurrent condition lasting longer than the overcurrent timeout.  • A device over temperature condition.  If the fault is detected in the main outputs, FAULTC must be reset by toggling the ONC input. If the fault is in the auxiliary output, FAULTC must be reset by toggling both ONC and AUXONC. For the autorestart version, FAULTC is reset when the part initiates the next power-on cycle.
26	PWRGDC	Open-Drain Power-Good Output. PWRGDC goes low 160ms after all outputs of slot C reach their final value and the power MOSFETs are fully enhanced.
27,28	3.3AUXOC	Slot C 3.3V Auxiliary Power-Supply Output
29, 30, 31, 69, 70, 71	3.3VAUXIN	3.3V Auxiliary Supply Input. 3.3VAUXIN is the input to a charge pump that drives the internal MOSFETs connecting 3.3AUXIN to 3.3AUXO V <sub>3.3AUXIN</sub> is also used to power the internal control logic and analog references of the MAX5959/MAX5960 and must always be connected to a supply between 3V and 3.6V. Bypass 3.3AUXIN with at least a 0.1µF capacitor to GND.
32, 33	3.3AUXOD	Slot D 3.3V Auxiliary Power-Supply Output
34, 66	PGND	Power Ground. Connect externally to GND.
35	PWRGDD	Open-Drain Power-Good Output. PWRGDD goes low 160ms after all outputs of slot D reach their final value and the power MOSFETs are fully enhanced.
36	FAULTD	Open-Drain Fault Output Signal. FAULTD latches active-low whenever slot D outputs are shut down due to a fault. A fault is either of:  • An overcurrent condition lasting longer than the overcurrent timeout.  • A device over temperature condition.  If the fault is detected in the main outputs, FAULTD must be reset by toggling the OND input. If the fault is in the auxiliary output, FAULTD must be reset by toggling both OND and AUXOND. For the autorestart version, FAULTD is reset when the part initiates the next power-on cycle.
37	TIM	Overcurrent Timeout Programming Input. Connect a resistor between $500\Omega$ and $500k\Omega$ from TIM to GND to program $t_{FAULT}$ . Leave TIM unconnected for a default timeout of 11ms.
38	PORADJ	Power-On-Reset Programming Input. Connect a resistor between $500\Omega$ and $500k\Omega$ from PORADJ to GND to program the POR timing. Leave unconnected for a default value of 160ms. Connect PORADJ to GND to completely skip the POR time delay for $\overline{\text{PWRGD}}$ assertion.
39	GND	Ground
40	12VIN	12V Supply Input. V <sub>12VIN</sub> drives the gates of the MOSFETs connected to 3.3G 12VIN powers an internal charge pump that drives the gates of the MOSFETs connected to 12G Bypass 12VIN with a 1µF capacitor to GND. See the <i>Typical Application Circuit</i> and <i>Input Transients</i> section.
41	12GD	Slot D 12V Gate-Drive Output. Connect 12GD to the gate of slot D's 12V MOSFET. At power-up, V <sub>12GD</sub> is raised to the internal charge-pump voltage level by a constant current.
42	12SD-	Slot D 12V Negative Current-Sense Input. Connect 12SD- to the negative side of the current-sense resistor using the Kelvin-sensing technique to ensure accurate current sensing.

## Pin Description (continued)

PIN	NAME	FUNCTION
43	12SD+	Slot D 12V Positive Current-Sense Input. Connect the positive terminal of the current-sense resistor to 12SD+ using the Kelvin-sensing technique to ensure accurate current sensing.
44	3.3GD	Slot D 3.3V Gate-Drive Output. Connect 3.3GD to the gate of slot D's 3.3V MOSFET. At power-up, V <sub>3.3GD</sub> is charged to 5.5V above the 3.3V supply by a constant current derived from V <sub>12VIN</sub> . V <sub>3.3GD</sub> 's rise time is determined by the external gate capacitance.
45	3.3SD-	Slot D 3.3V Negative Current-Sense Input. Connect to the negative side of the sense resistor using the Kelvin-sensing technique to ensure accurate current sensing.
46	3.3SD+	Slot D 3.3V Positive Current-Sense Input. Connect the positive side of the current-sense resistor to 3.3SD+ using the Kelvin-sensing technique to ensure accurate current sensing.
47	PRES-DETB	Slot B Present-Detect Input. PRES-DETB accepts inputs from PRSNT#_# on a PCIe connector. PRES-DETB has an internal 50kΩ pullup to 3.3VAUXIN. When PRES-DETB is low, the outputs follow the command from ONB and AUXONB after a 4ms debounced time. When PRES-DETB goes from low to high, all outputs of the respective slot shut down with no delay.
48	FONB	Slot B Forced-On Input. $\overline{\text{FONB}}$ has a 50k $\Omega$ internal pullup to 3.3VAUXIN. A logic-low on $\overline{\text{FONB}}$ turns on all slot B outputs as long as the power inputs are within their operating range, regardless of the status of the other input signals. Leave $\overline{\text{FONB}}$ unconnected for normal operation.
49	PRES-DETA	Slot A Present-Detect Input. PRES-DETA accepts inputs from PRSNT#_# on a PCIe connector.  PRES-DETA has an internal 50kΩ pullup to 3.3VAUXIN. When PRES-DETA is low, the outputs follow the command from ONA and AUXONA after a 4ms debounced time. When PRES-DETA goes from low to high, all outputs of the respective slot shut down with no delay.
50	FONA	Slot A Forced-On Input. $\overline{\text{FONA}}$ has a 50k $\Omega$ internal pullup to 3.3VAUXIN. A logic-low on $\overline{\text{FONA}}$ turns on all slot A outputs as long as the power inputs are within their operating range, regardless of the status of the other input signals. Leave $\overline{\text{FONA}}$ unconnected for normal operation.
51	OUTPUT4	Digital Output 4. 4ms debounced digital output of INPUT4.
52	INPUT4	Digital Logic Gate Input 4
53	OUTPUT3	Digital Output 3. 4ms debounced digital output of INPUT3.
54	INPUT3	Digital Logic Gate Input 3
55	ONA	Slot A 12V and 3.3V Outputs Enable. A logic-high at ONA turns on the 12V and 3.3V outputs of slot A (see Table 2).
56	AUXONA	Slot A 3.3V Auxiliary Output Enable. A logic-high at AUXONA turns on the slot A auxiliary output.
57	12SA+	Slot A 12V Positive Current-Sense Input. Connect the positive terminal of the current-sense resistor to 12SA+ using the Kelvin-sensing technique to ensure accurate current sensing.
58	12SA-	Slot A 12V Negative Current-Sense Input. Connect 12SA- to the negative side of the current-sense resistor using the Kelvin-sensing technique to ensure accurate current sensing.
59	12GA	Slot A 12V Gate-Drive Output. Connect 12GA to the gate of slot A's 12V MOSFET. At power-up, V <sub>12GA</sub> is raised to the internal charge-pump voltage level by a constant current.
60	3.3SA+	Slot A 3.3V Positive Current-Sense Input. Connect the positive side of the current-sense resistor to 3.3SA+ using the Kelvin-sensing technique to ensure accurate current sensing.
62	3.3SA-	Slot A 3.3V Negative Current-Sense Input. Connect to the negative side of the sense resistor using the Kelvin-sensing technique to ensure accurate current sensing.

### **Pin Description (continued)**

PIN	NAME	FUNCTION
63	3.3GA	Slot A 3.3V Gate-Drive Output. Connect 3.3GA to the gate of slot A's 3.3V MOSFET. At power-up, V3.3GA is charged to 5.5V above the 3.3V supply by a constant current derived from V <sub>12VIN</sub> . V <sub>3.3GA</sub> 's rise time is determined by the external gate capacitance.
64	FAULTA	Open-Drain Fault Output Signal. FAULTA latches active low whenever slot A outputs are shut down due to a fault. A fault is either of:  • An overcurrent condition lasting longer than the overcurrent timeout.  • A device over temperature condition.  If the fault is detected in the main outputs, FAULTA must be reset by toggling the ONA input. If the fault is in the auxiliary output, FAULTA must be reset by toggling both ONA and AUXONA. For the autorestart version, FAULTA is reset when the part initiates the next power-on cycle.
65	PWRGDA	Open-Drain Power-Good Output. PWRGDA goes low 160ms after all outputs of slot A reach their final value and the power MOSFETs are fully enhanced.
67, 68	3.3AUXOA	Slot A 3.3V Auxiliary Power-Supply Output
72, 73	3.3AUXOB	Slot B 3.3V Auxiliary Power-Supply Output
74	PWRGDB	Open-Drain Power-Good Output. PWRGDB goes low 160ms after all outputs of slot B reach their final value and the power MOSFETs are fully enhanced.
75	FAULTB	Open-Drain Fault Output Signal. FAULTB latches active-low whenever slot B outputs are shut down due to a fault. A fault is either of:  • An overcurrent condition lasting longer than the overcurrent timeout.  • A device over temperature condition.  If the fault is detected in the main outputs, FAULTB must be reset by toggling the ONB input. If the fault is in the auxiliary output, FAULTB must be reset by toggling both ONB and AUXONB. For the autorestart version, FAULTB is reset when the part initiates the next power-on cycle.
76	3.3GB	Slot B 3.3V Gate-Drive Output. Connect 3.3GB to the gate of slot B's 3.3V MOSFET. At power-up, V <sub>3.3GB</sub> is charged to 5.5V above the 3.3V supply by a constant current derived from V <sub>12VIN</sub> . V <sub>3.3GB</sub> 's rise time is determined by the external gate capacitance.
77	3.3SB-	Slot B 3.3V Negative Current-Sense Input. Connect to the negative side of the sense resistor using the Kelvin-sensing technique to ensure accurate current sensing.
78	3.3SB+	Slot B 3.3V Positive Current-Sense Input. Connect the positive side of the current-sense resistor to 3.3SB+ using the Kelvin-sensing technique to ensure accurate current sensing.
79	12GB	Slot B 12V Gate-Drive Output. Connect 12GB to the gate of slot B's 12V MOSFET. At power-up, V <sub>12GB</sub> is raised to the internal charge-pump voltage level by a constant current.

## **Detailed Description**

The MAX5959/MAX5960 quad hot-plug controllers are designed for PCIe applications. The devices provide hot-plug control for 12V, 3.3V, and 3.3V auxiliary supplies for three PCIe slots. The MAX5959/MAX5960s' logic inputs/outputs allow interfacing directly with the system hot-plug-management controller or through an SMBus with an external I/O expander. An integrated debounced attention switch and present-detect signals are included to simplify system design (Figure 1).

The MAX5959/MAX5960 drive eight external n-channel MOSFETs to control the 12V and 3.3V main outputs. The 3.3V auxiliary outputs are controlled through internal 0.2 $\Omega$  n-channel MOSFETs. Internal charge pumps provide a gate drive for the 12V outputs while the gate drive of the 3.3V output is driven by the 12V input supply. The 3.3V auxiliary outputs are completely independent from the main outputs with their own charge pumps.

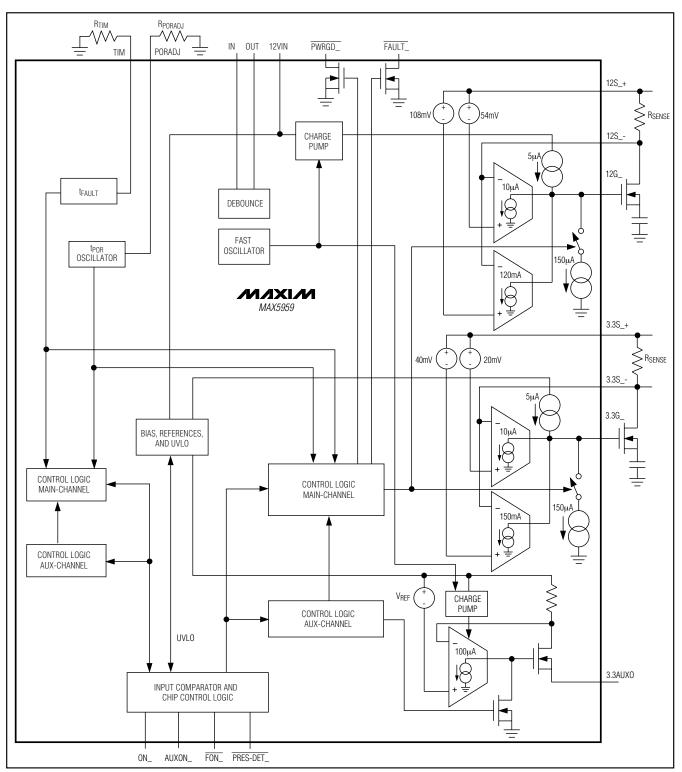


Figure 1. Single-Channel Internal Block Diagram

At power-up, the MAX5959/MAX5960 keep all the external MOSFETs off until all supplies rise above their respective UVLO thresholds. These devices keep the internal MOSFETs off only until the 3.3VAUXIN supply rises above its UVLO threshold. Upon a turn-on command, the MAX5959/MAX5960 enhance the external and internal MOSFETs slowly with a constant gate current to limit the power-supply inrush current. The MAX5959/MAX5960 actively limit the current of all outputs at all times and shut down the corresponding channel if an overcurrent condition persists for longer than a resistor-programmable overcurrent timeout (see the Fault Management section). Thermal protection circuitry also shuts down all outputs if the die temperature exceeds +150°C. After an overcurrent or overtemperature fault condition, the MAX5959/MAX5960 latch off or automatically restart depending on the version, after a restart time delay.

The power requirement for PCle connectors is defined by the PCle card specification and summarized in Table 1.

### Startup

The main supply outputs can become active only after all the following events have occurred:

- V<sub>3.3AUXIN</sub> is above its UVLO threshold.
- V<sub>12VIN</sub> and V<sub>3.3SA+</sub> are both above their UVLO threshold.
- ON\_ is driven high.
- PRES-DET\_ is low for more than 4ms.

The auxiliary supply output is made available only after the following events have occurred:

- V<sub>3</sub> 3AUXIN is above its UVLO threshold.
- AUXON is driven high.
- PRES-DET\_ is low for more than 4ms.

The FON\_ input overrides all other control signals and turns on the respective slot when driven low, as long as the UVLO thresholds have been reached. Table 2 summarizes the logic conditions required for startup. The auxiliary supply input powers the internal control logic and analog references of the MAX5959/MAX5960, so the main supplies cannot be enabled, if V3.3VAUXIN is not present. When an output is enabled, a programmable startup timer (tsu) begins to count the startup time duration.

The value of tsu is set to 2x the fault timeout period (tFAULT). RTIM externally connected from TIM to GND sets the duration of tFAULT.

Table 1. Power Requirement for PCIe Connectors

POWER FAIL	X1 CONNECTOR	X4/8 CONNECTOR	X16 CONNECTOR
3.3V			
Voltage Tolerance	±9% (max)	±9% (max)	±9% (max)
Supply Current	3.0A (max)	3.0A (max)	3.0A (max)
Capacitive Load	1000μF (max)	1000μF (max)	1000μF (max)
12V			
Voltage Tolerance	±8% (max)	±8% (max)	±8% (max)
Supply Current	0.5A (max)	2.1A (max)	5.5A (max)
Capacitive Load	300µF (max)	1000μF (max)	2000μF (max)
3.3V AUXILIARY			
Voltage Tolerance	±9% (max)	±9% (max)	±9% (max)
Supply Current, Wake Enabled	375mA (max)	375mA (max)	375mA (max)
Supply Current, Nonwake Enabled	20mA (max)	20mA (max)	20mA (max)
Capacitive Load	150µF (max)	150μF (max)	150µF (max)

**Table 2. Control Logic Truth Table** 

ON_	AUXON_	FON_	PRES-DET	12V_ AND 3.3V_ OUTPUTS	3.3VAUXO_AUXILIARY OUTPUTS
X	X	Low	X	On	On
X	X	High	High	Off	Off
Low	Low	High	Low*	Off	Off
High	Low	High	Low*	On	Off
Low	High	High	Low*	Off	On
High	High	High	Low*	On	On

<sup>\*</sup>PRES-DET\_ high-to-low transition has a 4ms delay (tDEG).

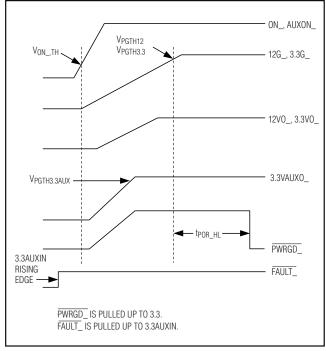


Figure 2. Power-Up Timing, No Fault

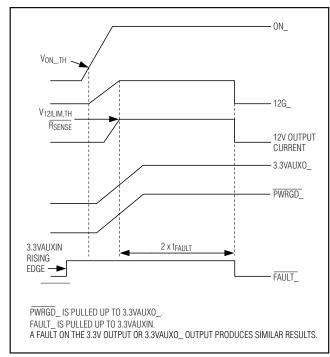


Figure 3. 12V Power-Up Timing (Turn-On into Output Overcurrent/Short Circuit)

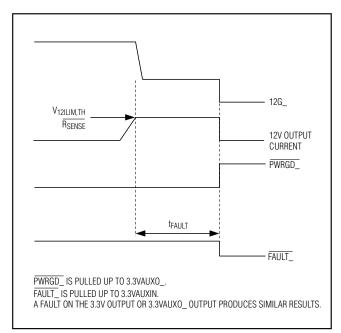


Figure 4. 12 Output Overcurrent/Short Circuit During Normal Operation

#### 12V and 3.3V Outputs Normal Operation

The MAX5959/MAX5960 monitor and actively limit the current of the 12V and 3.3V outputs after the startup period. Each output has its own overcurrent threshold. If any of the monitored output currents rise above the overcurrent threshold for a period tfault, FAULT asserts and the controller disengages both the 12V and 3.3V outputs for the particular slot (see the Fault Management section).

#### 3.3V Auxiliary Output Normal Operation

The auxiliary output current is internally monitored and actively limited to the maximum current-limit value. An overcurrent fault condition occurs when the output current exceeds the overcurrent threshold for longer than tfault. A fault on an auxiliary channel causes all supplies of the affected channel to be disabled after a programmable time period tfault.

A fault condition on a main channel ( $V_{12VIN}$  or  $V_{3.3VIN}$ ) causes all the channel's main outputs to shut down after the  $t_{FAULT}$  period and then either latch off or automatically restart after the  $t_{RESTART}$  ( $t_{RESTART} = 64 \times t_{FAUALT}$ ) period, depending on the device version. A

fault on any of the channel's main output does not affect the auxiliary channel (V<sub>3.3AUXIN</sub>).

### Power-Good (PWRGD\_)

Power-good (PWRGD\_) is an open-drain output that pulls low a time (tpOR\_HL) after all the outputs of the respective slot are fully on. All outputs are considered fully on when 3.3G\_ has risen to VpGTH3.3, 12G\_ has risen to VpGTH12, and V3.3AUXO\_is less than VpGTH3.3AUX. tpOR\_HL is adjustable from 2.4ms to 1.5s by connecting a resistor from PORADJ to GND. See the Setting the Power-On Reset and Timeout Period (tpOR\_HL) sections. Connect PORADJ to GND to completely skip the POR time delay for PWRGD\_ assertion.

#### **Thermal Shutdown**

When the die temperature goes above (TsD) +150°C, an overtemperature fault occurs and the MAX5959/MAX5960 shut down all outputs. The device waits for the junction temperature to decrease below TsD - hysteresis before entering fault management (see the *Fault Management* section):

### Fault Management

A fault occurs when an overcurrent or 12G\_ or 3.3G\_ below their power-good threshold lasts longer than trault or when the device experiences an overtemperature condition:

- A fault on a main output (12V or 3.3V) shuts down both main outputs of the respective slot. The 3.3V auxiliary is not affected.
- A fault on the 3.3V auxiliary output shuts down all three outputs of the respective slot.

The MAX5959A/MAX5960A automatically restart from a fault shutdown after the trestart period while the MAX5959L/MAX5960L latch off. If an overcurrent fault occurred on a main output, bring ON\_ low for at least trest (100µs) and high again to reset the fault and restart the outputs. If the overcurrent fault occurred on an auxiliary output or an overtemperature fault occurred, bring both ON\_ and AUXON\_ low for a minimum of trest to reset the fault. Toggle ON\_ or only AUXON\_ to reset the fault condition. If ON\_ and AUXON are toggled before trestart time counting has elapsed, the MAX5959L/MAX5960L store the information and restart when the delay is finished. The MAX5959A/MAX5960A (autoretry versions) restart all channels automatically after trestart.

# **Debounced Logic Gate** (INPUT\_ and OUTPUT\_)

INPUT1, INPUT2, and INPUT3 accept inputs from mechanical switches. The corresponding outputs are OUTPUT1, OUTPUT2, OUTPUT3, and OUTPUT4. OUTPUT\_ is debounced for 4ms. When INPUT\_ goes from high to low, OUTPUT\_ goes low immediately and stays low for at least 4ms. After the debounce time, OUTPUT\_ follows INPUT\_. If INPUT\_ goes from low to high, OUTPUT\_ goes high immediately and stays high for at least 4ms. After the debounce time, OUTPUT\_ follows INPUT\_. Figure 5 shows the timing diagram describing the INPUT /OUTPUT\_ debounced feature.

# Present-Detect and Forced-On Inputs (PRES-DET\_, FON\_)

PRES-DET\_ input detects the PRSNT\_# pin on a PCIe connector. When the card is plugged in, PRES-DET\_ goes low and allows the turn-on of the outputs of the respective slot after a 4ms debounced time. When the card is removed, an internal  $50k\Omega$  pullup resistor forces PRES-DET\_ high and the respective slot is shut down with no delay. PRES-DET\_ works in conjunction with ON\_ and AUXON\_ and only enables the device when ON\_ and AUXON\_ are high.

A logic-low on FON\_ forces the respective slot (main supplies and auxiliary) to turn on regardless of the status of the other logic inputs, provided the UVLO thresholds are exceeded on all the inputs.

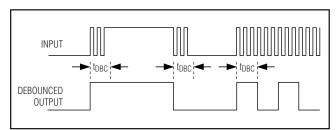


Figure 5. INPUT\_ AND OUTPUT\_ Debounced Feature

#### **Active Current Limits**

Active current limits are provided for all three outputs of the four slots (slot A, slot B, slot C, and slot D). Connect a current-sense resistor between 12S\_+ and 12S\_- to set the current limit for the 12V outputs. The current limit is set to 54mV / RSENSE12. Connect a current-sense resistor between 3.3S\_+ and 3.3S\_- to set the current limit for the 3.3V main outputs to 20mV / RSENSE3.3. For the auxiliary output (3.3VAUXO\_) the current limit is fixed at 450mA in the MAX5959 and 700mA in the case of the MAX5960.

When the voltage across Rsense12 or Rsense3.3 reaches the current-limit threshold voltage, the MAX5959/MAX5960 regulate the gate voltage to maintain the current-limit threshold voltage across the sense resistor. If the current limit lasts for tfault, then an overcurrent fault occurs. The MAX5959/MAX5960 shut down both the 12V and 3.3V outputs and assert the FAULT\_ output of the respective slot.

When the auxiliary output reaches the current limit 450mA (MAX5959) or 700mA (MAX5960) for longer than tFAULT, a fault occurs and the device shuts down all outputs and asserts FAULT of the respective slot.

#### **UVLO Threshold**

The UVLO thresholds prevent the internal auxiliary MOSFETs and the external main channel MOSFETs from turning on if V<sub>12VIN</sub>, V<sub>3.3VIN</sub>, and V<sub>3.3VAUXIN</sub> are not present. Internal comparators monitor the main supplies and the auxiliary supply and keep the gate-drive outputs (12GA, 12GB, 12GC, 12GD, 3.3GA, 3.3GB, 3.3GC, and 3.3GD) low until the supplies rise above their UVLO threshold. The 12V main supply is monitored at 12VIN and has a UVLO threshold of 10V. The 3.3V main supply is monitored at 3.3SA+ and has a UVLO threshold of 2.65V. The auxiliary supply is monitored at 3.3AUXIN and has a 2.65V UVLO threshold. For either main channels to operate, V3.3AUXIN must be above its UVLO threshold.

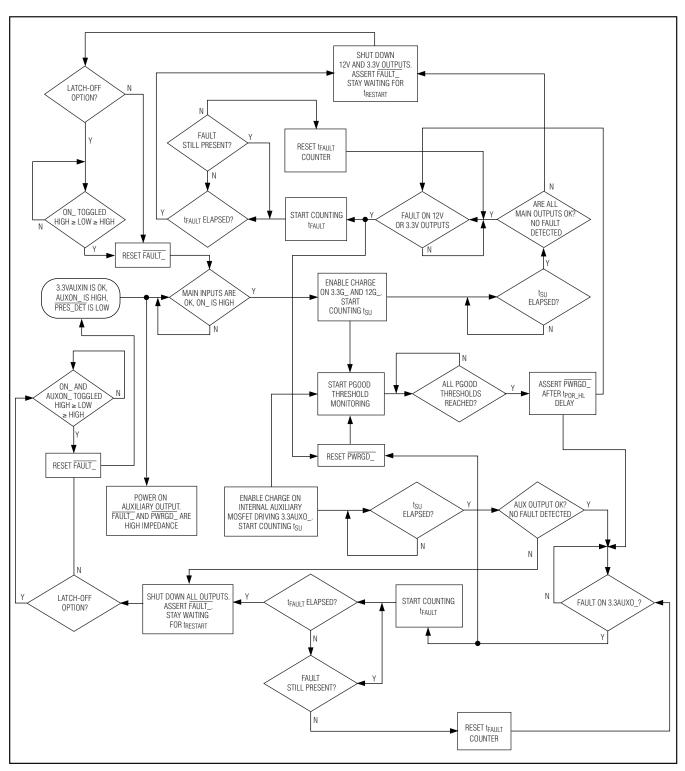


Figure 6. Fault Management Flowchart

20 \_\_\_\_\_\_\_/N|X|/V|

# External MOSFET Gate Drivers (12G\_ and 3.3G\_)

The gate drive for the external MOSFETs is provided at 12GA, 12GB, 12GC, 12GD, 3.3GA, 3.3GB, 3.3GC, and 3.3GD. 12G\_ is the gate drive for the 12V main supply and is boosted to 5.3V above V<sub>12VIN</sub> by its internal charge pump. During turn-on, 12G\_ sources 5µA into the external gate capacitance to control the turn-on time of the external MOSFET. During turn-off, 12G\_ sinks 150µA from the external gate capacitance to quickly turn off the external MOSFET. During short-circuit events, an internal 120mA current sink activates to rapidly bring the load current into the regulation limits.

3.3G\_ is the gate drive for the 3.3V main supply's MOS-FET and is driven to 5.5V above the 3.3V main supply. The power for 3.3G\_ is supplied from 12VIN and has no internal charge pump. During turn-on, 3.3G\_ sources 5µA into the external gate capacitance to control the turn-on time of the external MOSFET. During turn-off, 3.3G\_ sinks 150µA to quickly turn off the external MOSFET. During short-circuit events, an internal 120mA current sink activates to rapidly turn off the appropriate external MOSFET.

### **Auxiliary Supply (3.3VAUXIN)**

3.3VAUXIN provides power to the auxiliary outputs as well as the internal logic and references. The drains of the internal auxiliary MOSFETs connect to 3.3AUXIN through internal sense resistors and the sources connect to the auxiliary outputs (3.3VAUXO\_). Both MOSFETs have typical on-resistance of 0.2 $\Omega$ . Each channel's internal charge pump boosts the gate-drive voltage to fully turn on the internal n-channel MOSFETs. The auxiliary supplies have an internal current limit set to 450mA (MAX5959) or 700mA (MAX5960).

## Applications Information

### **Setting the Power-On Reset**

tfault is the time an overcurrent or overtemperature fault must remain for the MAX5959/MAX5960 to disable the main or auxiliary channels of a particular slot. Program the fault timeout period (tfault) by connecting a resistor (RTIM) from TIM to GND. tfault can be calculated by the following equation:

 $t_{FAULT} = (166 \text{ns} / \Omega) \times R_{TIM}$ 

The tFAULT programmed time duration must be chosen according to the total capacitance load connected to the 12G\_ and 3.3G\_ pins. To properly power up the main supply outputs, the following constraints need to be taken:

tsu ≥ (Vgate x Cload) / Ichg

where  $t_{SU} = 2 \times t_{FAULT}$  and where:

- ICHG = 5μA.
- VGATE = 4.8V +V<sub>12VIN</sub> for 12G\_ and VGATE = 6.8V +V<sub>3.3VIN</sub> for 3.3G\_.
- CLOAD is the total capacitance load at the gate.

Maximum and minimum values for RTIM are  $500\Omega$  and  $500k\Omega$ , respectively. Leave TIM floating for a default trault of 10ms.

### Timeout Period (tpor HL)

tpor\_HL is the time from when the gate voltages of all outputs of a slot reach their power-good threshold to when PWRGD\_ pulls low. Program the POR timeout period (tpor) by connecting a resistor (Rporadu) from PORADJ to GND. tpor\_HL can be calculated by the following equation:

$$tpor_{HL} = (2.5\mu s / \Omega) \times Rporadj$$

Maximum and minimum values for RPORADJ are  $500\Omega$  and  $500k\Omega$ , respectively. Leave PORADJ floating for a default tpor of 150ms. Connect PORADJ to GND in order to completely skip the power-on delay time prior to the  $\overline{PWRGD}$  assertion.

### **Component Selection**

Select the external n-channel MOSFET according to the applications current requirement. Limit the switch power dissipation by choosing a MOSFET with an RDS\_ON low enough to have a minimum voltage drop at full load. High RDS\_ON causes larger output ripple if there are pulsed loads. High RDS\_ON can also trigger an external undervoltage fault at full load. Determine the MOSFET's power-rating requirement to accommodate a short-circuit condition on the board during startup. Table 3 lists the MOSFETs and sense resistor manufacturers.

### **Table 3. Component Manufacturers**

COMPONENT	MANUFACTURER	PHONE	WEBSITE
Sense Resistor	Vishay-Date	402-564-3131	www.vishay.com
Sense nesision	IRC	704-264-8861	www.irctt.com
	Fairchild	888-522-5372	www.fairchildsemi.com
MOCEET	International Rectifier	310-322-3331	www.irf.com
MOSFETs	Motorola	602-244-3576	www.mot-sps.com/ppd
	Vishay-Siliconix	_	www.vishay.com

### **Additional External Gate Capacitance**

External capacitance can be added from the gate of the external MOSFETs to GND to slow down the dV/dt of the 12V and 3.3V outputs. The maximum gate capacitance load at 12G\_ and 3.3G\_ must be consistent with the conditions described in the *Setting the Power-On Reset* section.

### **Maximum Load Capacitance**

Large capacitive loads at the 12V output, the 3.3V output, and the 3.3V auxiliary output can cause a problem when inserting discharged PCI cards into live backplanes. A fault occurs if the time needed to charge the capacitance of the board is greater than the typical startup time (2 x tFAULT). The MAX5959/MAX5960 withstand large capacitive loads due to their adjustable startup times and adjustable current-limit thresholds. Calculate the maximum load capacitance as follows:

CLOAD < (tSU x ILIM) / VOUT

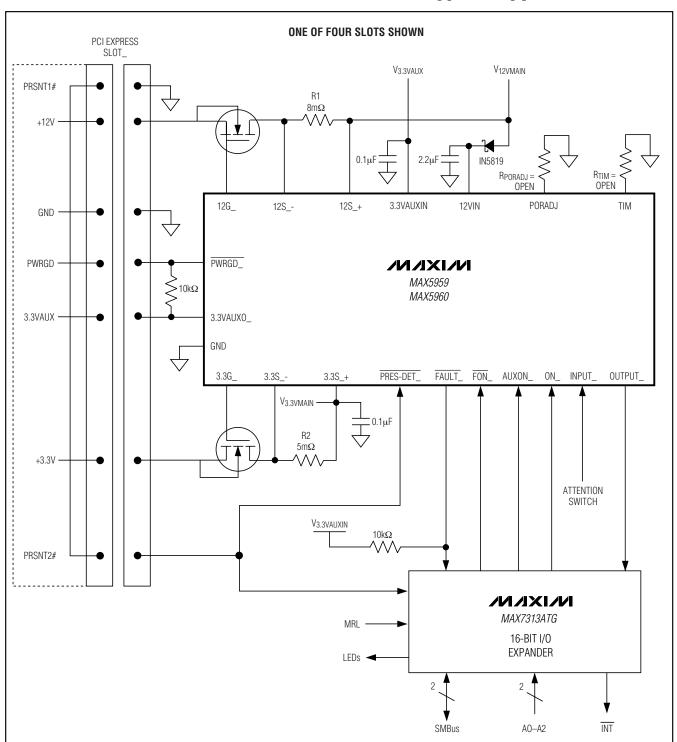
V<sub>OUT</sub> is either the 3.3V output, the 12V output, or the 3.3V auxiliary output for slot A, slot B, slot C, or slot D.

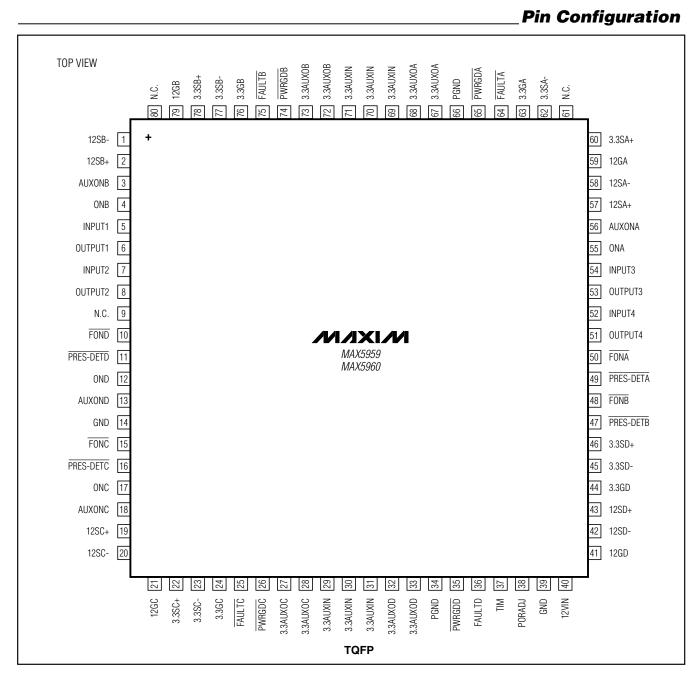
### Input Transients

The 12V input (12VIN), the 3.3V input (3.3SA+), and the 3.3V auxiliary (3.3AUXIN) must be above their UVLO thresholds before startup can occur. Input transients can cause the input voltage to sag below the UVLO threshold. The MAX5959/MAX5960 reject transients on the input supplies that are shorter than 4µs (typ).

Because some load fault conditions can cause voltage transients to propagate to the supply inputs with duration of greater than 4µs, it is recommended that a small Schottky diode be placed in series with the 12VIN pin connection, upstream of the 1µF bypass capacitor. This provides a hold-up supply that will prevent the 12VIN input from dropping below V12UVLO during severe transients. See the *Typical Application Circuit*.

## **Typical Application Circuit**





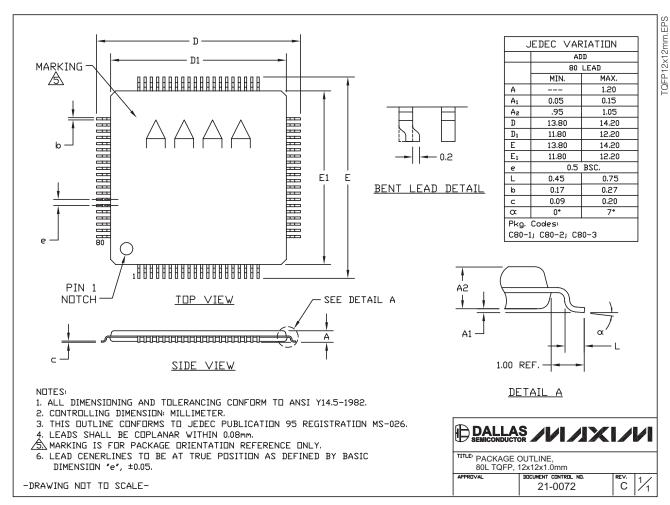
**Chip Information** 

PROCESS: BiCMOS

24 \_\_\_\_\_\_\_/N/XI/M

### Package Information

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to <a href="https://www.maxim-ic.com/packages">www.maxim-ic.com/packages</a>.)



Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.

## **X-ON Electronics**

Largest Supplier of Electrical and Electronic Components

Click to view similar products for Hot Swap Voltage Controllers category:

Click to view products by Maxim manufacturer:

Other Similar products are found below:

MP5010BDQ-LF-Z MP5022AGQV-P MP5010BDQ-LF-P MP5014GJ-P MP5018GD-P MAX5933CESA MAX5925CEUB MAX5909UEE

MAX5907EEE MAX5902AAEUT MAX5972BETE+ MAX5925DEUB+T MAX15091AETI+ MAX5956AUEE+ MAX5939BESA+

MAX5929CEEG+ MAX5929AEEG+ MP5010ADQ-LF-Z STEF05LJR STEF05WPUR MAX14693ATP+T STEF12WPUR

MAX5971BETI+T NCP81295MNTXG LTC4227IUFD-1#PBF LTC4235CUFD-1#PBF LT1641-1IS8#TRPBF MAX5910ESA+T

STEF05LPUR STEF05LAPUR MAX15093GWL+T ADM1275-3ARQZ-R7 TPS259807ONRGER TPS259802ONRGER LT4256
2IS8#TRPBF TPS25944ARVCR LPTM21-1AFTG237C NCP1094GEVB TPS25942LRVCR ADM1270ACPZ-R7 UCC2921DTR

ADM4210-2AUJZ-RL7 ADM1170-2AUJZ-RL7 ADM1075-2ACPZ MAX5988BETP+ MAX5981AETE+ MAX5978ETJ+

MAX5977AETP+ MAX5969DETE+ MAX5914AEMH+