

ZXRE160\_H5 (SC70/SOT353) and



# **ZXRE160**

#### 0.6V ENHANCED ADJUSTABLE PRECISION SHUNT REGULATOR

Pin Assignments

## Description

The ZXRE160 is a 5-terminal adjustable shunt regulator offering excellent temperature stability and output handling capability. This device offers an enhancement to the ZXRE060 part for use in a comparator mode applications.

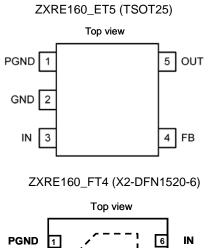
In shunt regulator mode, the ZXRE160 simplifies the design of isolated low voltage DC-DC regulators. With its low 0.6V FB pin, it can control the regulation of rails as low 0.6V. This makes the part ideal for state of the art microprocessor, DSP and PLD core voltage POL converters.

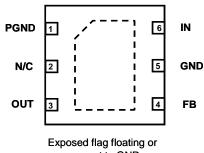
The device open-collector output can operate from 0.2V to 18V and regulated output voltage can be set by selection of two external divider resistors. Separating the input from the open collector output enables the ZXRE160 to be used to make low-cost low drop-out regulators operating at low input voltages.

The ZXRE160 is available in two grades with initial tolerances of 0.5% and 1% for the A and standard grades respectively. It is available in space saving low profile 5 pin SC70/SOT353, thin TSOT25 and very small DFN1520 packages.

#### Features

- Low reference voltage ( $V_{FB} = 0.6V$ )
- -40°C to +125°C temperature range
- Reference voltage tolerance at +25°C
- 0.5% ZXRE160A
- 1% ZXRE160
- Typical temperature drift
- <4 mV (0°C to +70°C)
- <6 mV (-40°C to +85°C)
- <12mV (-40°C to +125°C)
- 0.2V to 18V open-collector output
- High power supply rejection
- (>45dB at 300kHz)
- Totally Lead-Free & Fully RoHS Compliant (Notes 1 & 2)
- Halogen and Antimony Free. "Green" Device (Note 3)





#### connect to GND

#### Applications

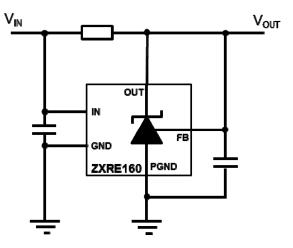
- Isolated DC-DC converters
- Core voltage POL
- Low Voltage Low-Dropout linear regulators
- Shunt regulators
- Adjustable voltage reference

- 1. No purposely added lead. Fully EU Directive 2002/95/EC (RoHS) & 2011/65/EU (RoHS 2) compliant. Notes:
  - 2. See http://www.diodes.com for more information about Diodes Incorporated's definitions of Halogen- and Antimony-free, "Green" and Lead-free.
  - 3. Halogen- and Antimony-free "Green" products are defined as those which contain <900ppm bromine, <900ppm chlorine (<1500ppm total Br + Cl) and <1000ppm antimony compounds.





## **Typical Applications Circuit**



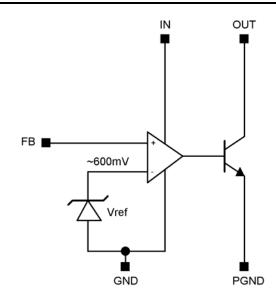
## **Pin Descriptions**

Pin Name Pin Number					
	SC70/ SOT353, TSOT25	X2-DFN1520-6	Function		
PGND	1	1	Power Ground: Ground return for emitter of output transistor: Connect PGND and GND together.		
_	—	2	No connection		
OUT	5	3	Output: Connect a capacitor close to device between OUT and GND for closed loop stability. See the <i>Applications Information</i> section.		
FB	4	4	Feedback Input. Threshold voltage 600mV nominal.		
GND	2	5	Analog Ground: Ground return for reference and amplifier: Connect GND and PGND together.		
IN	3	6	Supply Input: Connect a 0.1µF ceramic capacitor close to the device from IN to GND.		
_	_	Flag	Floating or connect to GND		





#### **Functional Block Diagram**



The ZXRE160 differs from most other shunt regulators in that it has separate input and output pins and a low voltage reference. This enables it to regulate rails down to 600mV and makes the part ideal for isolated power supply applications that use opto-couplers in the feedback loop and where the open-collector output is required to operate down to voltages as low as 200mV.

The wide input voltage range of 2V to 18V and output voltage range of 0.2V to 18V enables the ZXRE160 to be powered from an auxiliary rail, while controlling a master rail which is above the

auxiliary rail voltage, or below the minimum  $V_{\rm IN}$  voltage. This allows it to operate as a low-dropout voltage regulator for microprocessor/DSP/PLD cores.

As with other shunt regulators (and shunt references), the ZXRE160 compares its internal amplifier FB pin to a high accuracy internal reference; if FB is below the reference then OUT turns off, but if FB is above the reference then OUT sinks current – up to a maximum of 15mA.





#### Absolute Maximum Ratings (Voltages to GND, @TA = +25°C, unless otherwise specified.)

Symbol	Parameter	Rating	Unit
VIN	IN Voltage relative to GND	20	V
Vout	OUT Voltage relative to GND	20	V
V <sub>FB</sub>	FB Voltage relative to GND	20	V
PGND	PGND Voltage relative to GND	-0.3 to +0.3	V
lout	OUT Pin Current	20	mA
TJ	Operating Junction Temperature	-40 to 150	°C
T <sub>ST</sub>	Storage Temperature	55 to 150	°C

These are stress ratings only. Operation outside the absolute maximum ratings may cause device failure. Operation at the absolute maximum rating for extended periods may reduce device reliability.

Semiconductor devices are ESD sensitive and may be damaged by exposure to ESD events. Suitable ESD precautions should be taken when handling and transporting these devices.

## **Package Thermal Data**

Package	θ <sub>JA</sub>	P <sub>DIS</sub> T <sub>A</sub> = 25°C, T <sub>J</sub> = 150°C
SC70/SOT353	400°C/W	310mW
TSOT25	250°C/W	500mW
X2-DFN1520-6	TBD	TBD

## Recommended Operating Conditions (@T<sub>A</sub> = +25°C, unless otherwise specified.)

Symbol	Parameter	Min	Max	Units
VIN	IN Voltage Range (0 to +125°C)	2	18	
VIN	IN Voltage Range (-40°C to 0°C)	2.2	18	V
Vout	OUT Voltage Range	0.2	18	
I <sub>OUT</sub>	OUT Pin Current	0.3	15	mA
T <sub>A</sub>	Operating Ambient Temperature Range	-40	+125	°C





## Electrical Characteristics (@T<sub>A</sub> = +25°C, V<sub>DD</sub> = 3V, unless otherwise specified.)

 $T_A = +25^{\circ}C$ ,  $V_{IN} = 3.3V$ ,  $V_{OUT} = V_{FB}$ ,  $I_{OUT} = 5mA$ , unless otherwise specified.) (Note 4)

Symbol	Parameter		Conditions		Min	Тур	Max	Units
				ZXRE160A	0.597	0.6	0.603	
				ZXRE160	0.594	0.6	0.606	
		$T_{A} = 0^{\circ}C + 25^{\circ}C$	T. 0°C to 195°C		0.595		0.605	ļ
		$T_A = 0^{\circ}C \text{ to } +85^{\circ}C$		ZXRE160	0.592		0.608	
$V_{FB}$	Feedback voltage	T <sub>A</sub> = -40°C to +85°	°C	ZXRE160A	0.594		0.606	V
		14 - 40 0 10 103	0	ZXRE160	0.591		0.609	
		T <sub>A</sub> = -40°C to +125	5°C	ZXRE160A	0.593		0.607	
		14 - 40 0 10 1120		ZXRE160	0.590		0.610	
FB <sub>LOAD</sub>	Feedback pin load	$I_{OUT} = 1$ to 15mA				3.8	6	mV
1 DLOAD	regulation			T <sub>A</sub> = -40 to +125°C			10	
FB <sub>LINE</sub>	Feedback pin line	V <sub>IN</sub> = 2V to 18V				0.3	1	mV
	regulation	V <sub>IN</sub> = 2.2V to 18V		T <sub>A</sub> = -40 to +125°C			1.5	mv
	Output voltage	$V_{OUT} = 0.2V$ to 18V	<i>ι</i> ,				1	
FB <sub>OVR</sub> regulation		I <sub>OUT</sub> = 1mA (Ref. Figure 1)		T <sub>A</sub> = -40 to +125°C			1.5	mV
						-45		nA
I <sub>FB</sub>	FB input bias current	V <sub>IN</sub> = 18V	T <sub>A</sub> = -40 to +125°C	-200		0		
				V <sub>FB</sub> = 0.7V	-50		50	
		$V_{IN} = 2V$ to $18V$				0.35	0.7	<b>m</b> (
		V <sub>IN</sub> = 2.2V to 18V	I <sub>OUT</sub> = 0.3mA	T <sub>A</sub> = -40 to +125°C			1	mA
I <sub>IN</sub>	Input current	V <sub>IN</sub> = 2V to 18V	I <sub>OUT</sub> = 10mA			0.48	1	
		V <sub>IN</sub> = 2.2V to 18V		T <sub>A</sub> = -40 to +125°C			1.5	mA
		$V_{IN} = 18V, I_{OUT} = 0$	.3mA	V <sub>FB</sub> = 0.7V			3	
		V <sub>IN</sub> = 18V,					0.1	,
I <sub>OUT(LK)</sub>	OUT leakage current	DUT leakage current V <sub>OUT</sub> = 18V, V <sub>FB</sub> =0V		T <sub>A</sub> = +125°C			1	μA
_	Dynamic Output	$I_{OUT} = 1$ to 15mA				0.25	0.4	
Z <sub>OUT</sub>	Impedance	f < 1kHz		T <sub>A</sub> = -40 to +125°C			0.6	Ω
PSRR	Power supply rejection ratio	F = 300kHz V <sub>AC</sub> = 0.3V <sub>PP</sub>				>45		dB
BW	Amplifier Unity Gain Frequency	Ref: Figure 2		1		600		kHz
G	Amplifier Transconductance					5000		mA/V

Note:

4. Production testing of the device is performed at +25°C. Functional operation of the device and parameters specified over the operating temperature range are guaranteed by design, characterization and process control.

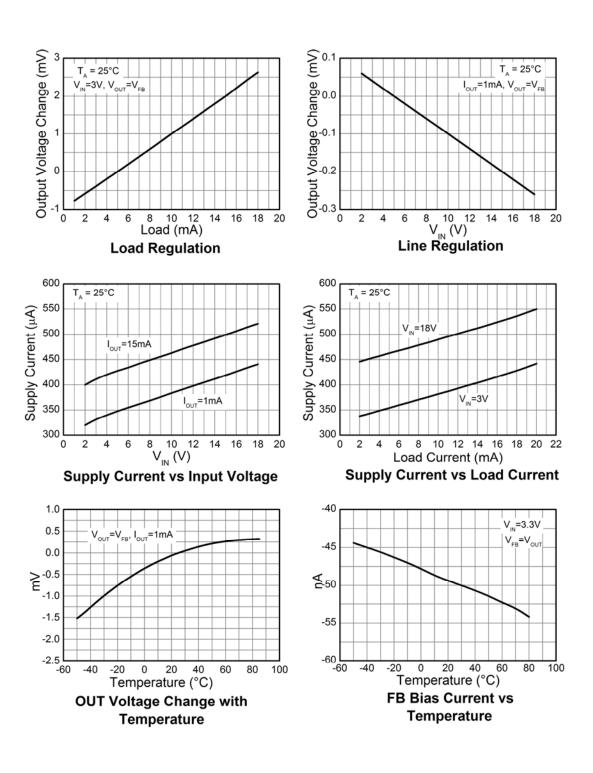






## **Typical Characteristics**

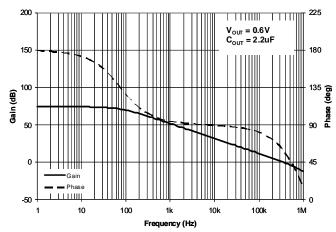
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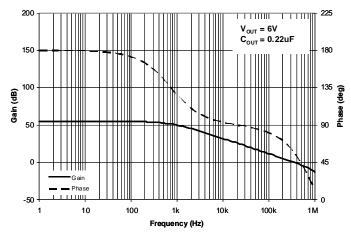




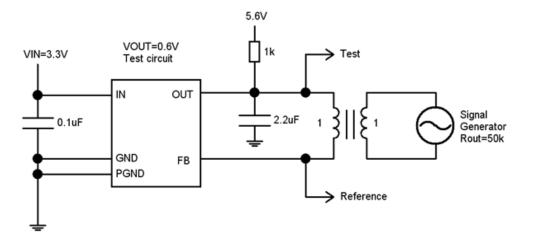
## Typical Operating Characteristics



Gain and Phase vs Frequency,  $V_{OUT}$ =0.6V



Gain and Phase vs Frequency,  $V_{OUT}$ =6V



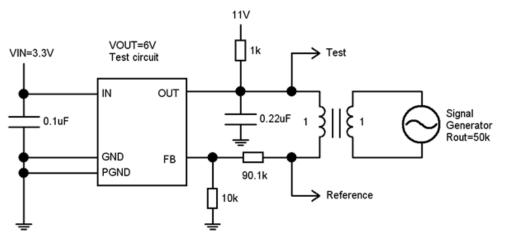


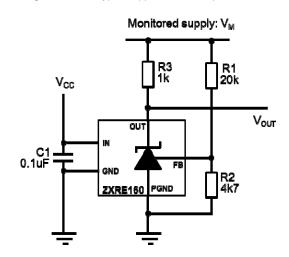
Figure 2. Test Circuits for Gain and Phase Plots





#### **Application Information**

The following show some typical application examples for the ZXRE160.



#### Figure 3. 15V Supply Monitor

Figure 3 shows a typical configuration for the ZXRE160 in comparator mode.

Here the comparator switches low when:

$$V_{M} \geq \frac{V_{FB}(R_{1} + R_{2})}{R_{2}}$$

Alternative values of R1, R2 may be used to provide different threshold voltages. R3 can also be adjusted to set the bias current for different values of  $V_M$ . R2 should be kept as low as possible to minimize errors due to the bias current of the FB pin.

This circuit has no hysteresis, so a small capacitor of approx.4.7nF between FB and GND is recommended to provide cleaner transitions at the output.

In shunt regulator mode it is necessary to include the compensation capacitor C2 to guarantee stability. C2 may range in value from  $0.1\mu$ F to  $10\mu$ F depending on the application. The minimum value of C2 can be determined from the following equation (resistor values are in k $\Omega$ ):

$$C2_{MIN} \ge \frac{R_2}{R_3(R_1 + R_2)} \mu F$$

Both C1 and C2 should be as close to the ZXRE160 as possible and connected to it with the shortest possible track. In the case of Figure 10 and Figure 11, it means the opto-coupler will have to be carefully positioned to enable this.

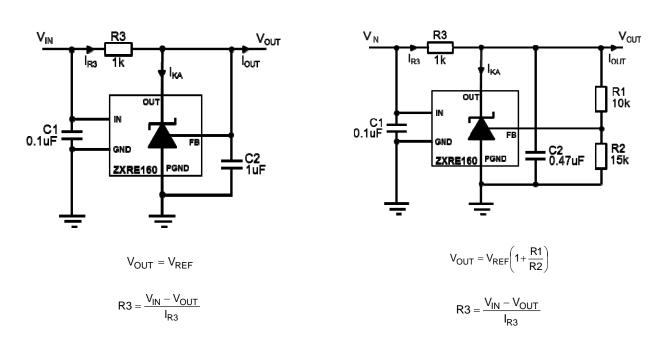


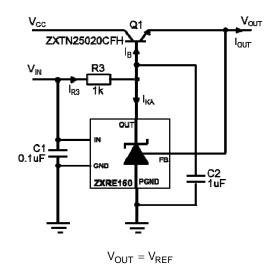
Figure 4. 0.6V Shunt Regulator

Figure 5. 1.0V Shunt Regulator





## Application Information (cont.)



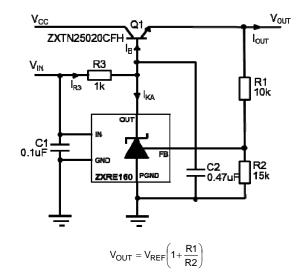


Figure 6. 0.6V Series LDO Regulator

Figure 7. 1.0V Series LDO Regulator

#### Design guide:

- 1. Determine  $I_{OUT}$  and choose a suitable transistor taking power dissipation into consideration.
- 2. Determine I<sub>B</sub> from I<sub>B</sub> =  $\frac{I_{OUT(max)}}{(h_{FE(min)} + 1)}$
- 3. Determine  $I_{R3}$  from  $I_{R3} \ge I_B + I_{KA(min)}$ . The design of the ZXRE160 effectively means there is no  $I_{KA(min)}$  limitation as in conventional references. There is only an output leakage current which is a maximum of 1µA. Nevertheless, it is necessary to determine an  $I_{KA(min)}$  to ensure that the device operates within its linear range at all times.  $I_{KA(min)} \ge 10\mu$ A should be adequate for this.
- 4. Determine R3 from  $R3 = \frac{V_{IN} (V_{OUT} + V_{BE})}{I_{R3}}$ .

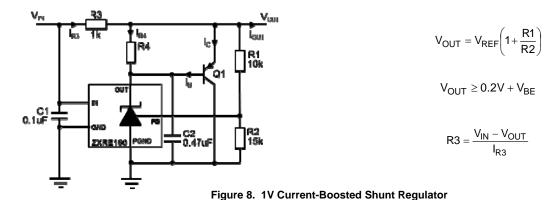
Although unlikely to be a problem, ensure that  $I_{R3} \le 15$  mA.







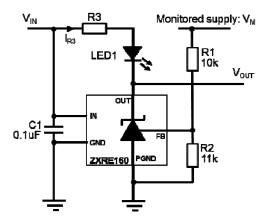
## Application Information (cont.)



#### Design guide

- 1. Determine I<sub>OUT</sub> and choose a suitable transistor taking power dissipation into consideration.
- 2. Determine I<sub>B</sub> from I<sub>B</sub> =  $\frac{I_{OUT(max)}}{(h_{FE(min)} + 1)}$
- 3. Determine  $I_{R3}$  from  $I_{R3} = I_{OUT(max)}$
- 4. Determine R3 from R3 =  $\frac{V_{IN} V_{OUT}}{I_{R3}}$
- It is best to let the ZXRE160 supply as much current as it can before bringing Q1 into conduction. Not only does this minimize the strain on Q1, it also guarantees the most stable operation. Choose a nominal value between 10mA and <15mA for this current, I<sub>R4</sub>.

Calculate R4 from R4 =  $\frac{V_{BE}}{I_{R4}}$ 



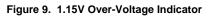
VOUT goes low and LED is lit when monitored supply

$$V_{M} > V_{REF} \left( 1 + \frac{R1}{R2} \right)$$

$$R3 = \frac{V_{IN} - (V_F + 0.2)}{I_{R3}}$$

 $15mA \ge I_{R3} \le I_{F(MAX)}$ 

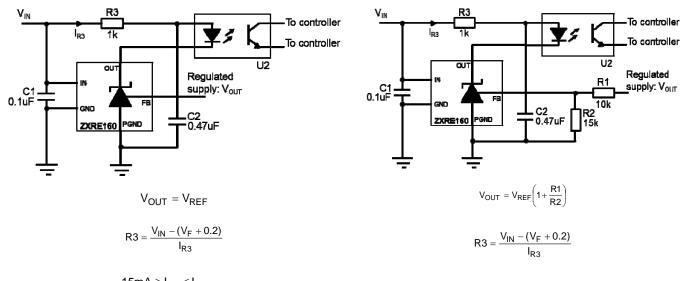
V<sub>F</sub> and I<sub>F</sub> are forward voltage drop and current of LED1.







## Application Information (cont.)



 $15mA \geq I_{R3} \leq I_{F(MAX)}$ 

Figure 10. Opto-Isolated 0.6V Shunt Regulator

 $15mA \geq I_{R3} \leq I_{F(MAX)}$ 

Figure 11. Opto-Isolated 1.0V Shunt Regulator

VF and IF are forward voltage drop and forward current respectively for the optocoupler LED

More applications information is available in the following publications which can be found on Diodes' web site.

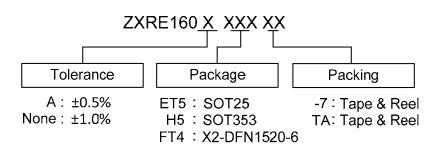
AN58 - Designing with Diodes' References – Shunt Regulation
AN59 - Designing with Diodes' References – Series Regulation
AN60 - Designing with Diodes' References – Fixed Regulators and Opto-Isolation
AN61 - Designing with Diodes' References – Extending the operating voltage range
AN62 - Designing with Diodes' References – Other Applications
AN63 - Designing with Diodes' References – ZXRE060 Low Voltage Regulator





#### Ordering Information

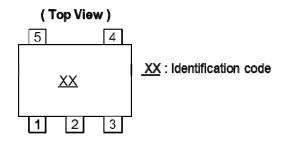
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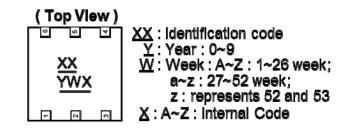
	Tol.	Part Number	Package	Identification Code	Reel Size	Tape Width	Quantity/Reel
		ZXRE160AET5TA	TSOT25	R8	7", 180mm	8mm	3000
Pb,	0.5%	ZXRE160AH5TA	SC70/SOT353	R9	7", 180mm	8mm	3000
Cead-free Green		ZXRE160AFT4-7	DFN1520H4-6	R8	7", 180mm	8mm	3000
		ZXRE160ET5TA	TSOT25	Z8	7", 180mm	8mm	3000
Pb,	1%	ZXRE160H5TA	SC70/SOT353	Z9	7", 180mm	8mm	3000
		ZXRE160FT4-7	X2-DFN1520-6	Z8	7", 180mm	8mm	3000

## **Marking Information**

1. TSOT25, SC70/SOT353



#### 2. X2-DFN1520-6



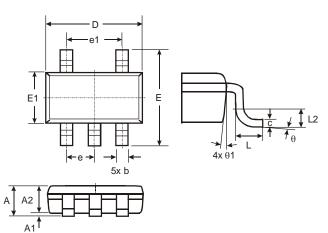






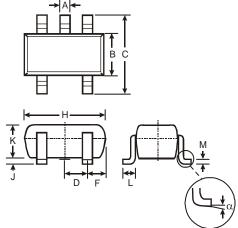
## Package Outline Dimensions (All dimensions in mm.)

TSOT25

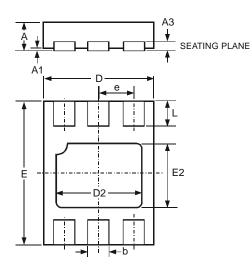


TSOT25					
Dim	Min	Max	Тур		
Α	-	1.00	-		
A1	0.01	0.10	-		
A2	0.84	0.90	-		
D	-	1	2.90		
Е	_	-	2.80		
E1	_	-	1.60		
b	0.30	0.45	_		
С	0.12	0.20	-		
е	-	-	0.95		
e1	-	-	1.90		
L	0.30	0.50			
L2	_	-	0.25		
θ	0°	8°	4°		
θ1	4°	12°	_		
All D	imensi	ons in	mm		

#### SC70/SOT353



#### X2-DFN1520-6



Α	0.10	0.30		
В	1.15	1.35		
С	2.00	2.20		
D	0.65 Typ			
F	0.40	0.45		
н	1.80	2.20		
J	0	0.10		
κ	0.90	1.00		
L	0.25	0.40		
М	0.10	0.22		
α	0°	8°		
All Dimensions in mm				

SOT353

Min

Max

Dim

)	X2-DFN1520-6						
Dim	Min	Max	Тур				
Α	1	0.40	-				
A1	0	0.05	-				
A3	1	-	0.13				
b	0.20	0.30	-				
D	1.45	1.575	-				
D2	1.00	1.20	-				
е	I	-	0.50				
Е	1.95	2.075	-				
E2	0.70	0.90	-				
L	0.25	0.35	_				
All D	imens	ions ir	n mm				

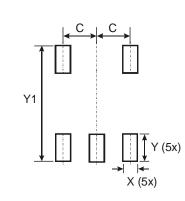






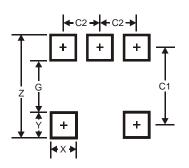
## **Suggested Pad Layout**

TSOT25



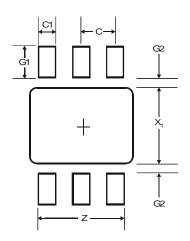
Dimensions	Value (in mm)
С	0.950
х	0.700
Y	1.000
Y1	3.199

SC70/SOT353



Dimensions	Value (in mm)
Z	2.5
G	1.3
Х	0.42
Y	0.6
C1	1.9
C2	0.65

#### X2-DFN1520-6



Dimensions	Value (in mm)
Z	1.25
G1	0.45
G2	0.15
X1	1.10
C	0.50
C1	0.25





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