# 4-Decade Current Sense Amplifier with Internal R<sub>SENSE</sub>

#### **General Description**

The MAX40016 is a very wide range current sense amplifier (CSA) with internal sense element that senses from less than 300µA to greater than 3A current range. The 4-decade sensed current functions with 1% (typical) gain error and offers three, multiplexed programmable output ranges in order to interface with 12-bit ADCs. Having an integrated sense element has the extra advantage that the entire current measuring path can be factory-trimmed, saving the user from having to calibrate independent sense resistors and CSAs. The MAX40016 drops a typical of 60mV at 3A from the voltage input to load output.

The MAX40016's integrated current-sensing element saves the space and cost of an external high-power, precision current sense resistor. The MAX40016 is offered in an ultra-tiny, 1.98mm x 1.31mm, 15-bump wafer-level package (WLP), further reducing board space.

The MAX40016 operates with a supply voltage from 2.5V to 5.5V. The device features a low-power mode in which the current-sensing element remains on, but the outputs are turned off to reduce the total supply current below  $10\mu A$  (max).

The MAX40016 also includes a committed on-board amplifier with an internal gain of 1.5V/V. The MAX40016 operates over the -40°C to +125°C temperature range.

#### **Applications**

- Mobile Devices
- RF Power Monitoring
- Portable Instruments

#### **Benefits and Features**

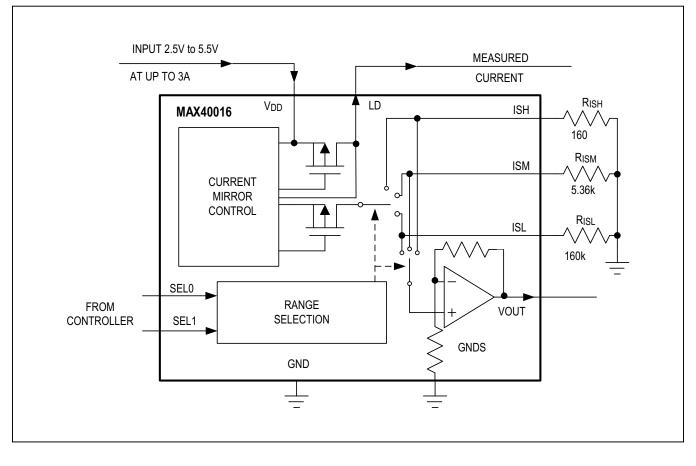
- Integrated Current Sense Element Saves The Space and Cost of Expensive Precision Sense Resistors
- 4-Decade Measurement Range
  Maintains Accuracy from < 300µA to > 3A
- Withstands Overloads to 4A
- Low Voltage Drop Across Sense Element
  - 60mV (Active Mode, 3A Load)
  - 35mV (Low Power Mode, 3A Load)
- Three Multiplexed Scaling Resistor Outputs Allow Full Dynamic Range while Interfaced to 12-bit ADCs
- +2.5V to +5.5V Input Supply Voltage Range
- Low Power Mode Reduces Supply Current to 10µA Max
- Space-Saving
  - Tiny 1.98mm x 1.3mm, 15-Bump WLP
  - 4mm x 4mm 16-Pin TQFN
- -40°C to +125°C Operating Temperature Range

Ordering Information appears at end of data sheet.



# 4-Decade Current Sense Amplifier with Internal R<sub>SENSE</sub>

### Simplified Block Diagram



# 4-Decade Current Sense Amplifier with Internal R<sub>SENSE</sub>

#### **Absolute Maximum Ratings**

V <sub>DD</sub> to GND	0.3V to +6V
GND to GNDS	0.3V to +0.3V
SEL0, SEL1, ISL, ISM, ISH, VOUT to	GND0.3V to V <sub>DD</sub> +0.3V
V <sub>DD</sub> to LD	0.3V to 0.3V
LD to GND	V <sub>DD</sub> - 0.3V to V <sub>DD</sub> + 0.3V
Maximum Current	
(All pins except V <sub>DD</sub> , LD, continuo	us)20mA
Current from V <sub>DD</sub> to LD (Continuous)	4A

Continuous Power Dissipation ( $T_A = +70^{\circ}C$ )
15-Bump WLP (derate 14.39mW/°C above +70°C) 1151.2mW
16-Pin TQFN (derate 25mW/°C above +70°C)2000mW
Operating Temperature Range40°C to +125°C
Junction Temperature+150°C
Storage Temperature Range65°C to +150°C
Soldering Temperature (reflow)+260°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### **Package Information**

#### **15 WLP**

PACKAGE CODE	N151B1+1				
Outline Number	21-100213				
Land Pattern Number	Refer to Application Note 1891				
THERMAL RESISTANCE, MULTI-LAYER BOAR	THERMAL RESISTANCE, MULTI-LAYER BOARD:				
Junction to Ambient ( $\theta_{JA}$ )	69.5°C/W				
Junction to Case $(\theta_{JC})$	N/A				

#### **16 TQFN**

PACKAGE CODE	T1644+4
Outline Number	<u>21-0139</u>
Land Pattern Number	90-0070
THERMAL RESISTANCE, MULTI-LAYER BOAR	D:
Junction to Ambient $(\theta_{JA})$	40°C/W
Junction to Case $(\theta_{JC})$	6°C/W

For the latest package outline information and land patterns (footprints), go to <u>www.maximintegrated.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to <a href="http://www.maximintegrated.com/thermal-tutorial">www.maximintegrated.com/thermal-tutorial</a>.

#### **Electrical Characteristics**

 $(V_{DD} = 3.6V, I_{LD} = 300mA, C_{LD} = 10\mu$ F, SEL0 =  $V_{DD}$ , SEL1 =  $V_{DD}$  (ISH range is selected),  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted. Typical values are at  $T_A = +25$ °C (Note 1))

PARAMETER	SYMBOL CONDITIONS		MIN	TYP	MAX	UNITS
CURRENT SENSING						
Supply Voltage	V <sub>DD</sub>	Guaranteed by PSRR	2.5		5.5	V
Supply Current (Active)	I <sub>DD</sub>	No I <sub>LD</sub> current, V <sub>ISX</sub> = 0V		0.8	1.2	mA
Supply Current (Low-Power Mode)	I <sub>DD_LP</sub>	Low-power mode (SEL0 = 0V, SEL1 = 0V), no I <sub>LD</sub> current, V <sub>ISX</sub> = 0V		5	10	μA
Power-Up Time		Measure at 50% of VOUT.		100		μs
Power Supply Rejection Ratio	PSRR	$\Delta$ Gain Error/ $\Delta$ V <sub>DD</sub> , measured at ISX (Note 2)	-0.6	+0.2	+0.6	%/V

# 4-Decade Current Sense Amplifier with Internal R<sub>SENSE</sub>

#### **Electrical Characteristics (continued)**

 $(V_{DD} = 3.6V, I_{LD} = 300mA, C_{LD} = 10\mu$ F, SEL0 =  $V_{DD}$ , SEL1 =  $V_{DD}$  (ISH range is selected),  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted. Typical values are at  $T_A = +25^{\circ}$ C (Note 1))

PARAMETER	SYMBOL	CON	DITIONS	MIN	TYP	MAX	UNITS
		Active mode,	-40°C < T <sub>A</sub> < +85°C		60	90	
		I <sub>LD</sub> = 3A (Note 6)	-40°C < T <sub>A</sub> < +125°C			100	
		Active mode,	-40°C < T <sub>A</sub> < +85°C		50	80	1
Voltage Drop (V <sub>DD</sub> to LD) (WLP)		I <sub>LD</sub> = 2A	-40°C < T <sub>A</sub> < +125°C			95	1
		Low power	-40°C < T <sub>A</sub> < +85°C		35	50	- mV
		mode, I <sub>LD</sub> = 3A (Note 6)	-40°C < T <sub>A</sub> < +125°C			55	1
		Low power	-40°C < T <sub>A</sub> < +85°C		23	35	1
		mode, I <sub>LD</sub> = 2A	-40°C < T <sub>A</sub> < +125°C			35	1
		Active mode,	-40°C < T <sub>A</sub> < +85°C		160	230	
		I <sub>LD</sub> = 3A (Note 6)	-40°C < T <sub>A</sub> < +125°C		160	250	1
		Active mode,	-40°C < T <sub>A</sub> < +85°C		120	180	]
		I <sub>LD</sub> = 2A	-40°C < T <sub>A</sub> < +125°C		120	200	]
Voltage Drop (V <sub>DD</sub> to LD) (TQFN)		Low power	-40°C < T <sub>A</sub> < +85°C		150	220	mV
		mode, I <sub>LD</sub> = 3A (Note 6)	-40°C < T <sub>A</sub> < +125°C		150	230	]
		Low power	-40°C < T <sub>A</sub> < +85°C		100	150	
		mode, I <sub>LD</sub> = 2A	-40°C < T <sub>A</sub> < +125°C		100	160	
Current Gain	Gl	I <sub>ISX</sub> /I <sub>LD</sub> , measure	ed at ISX		2		mA/A
		$\label{eq:RISX} \begin{array}{ c c c } R_{ISX} = 160\Omega, \\ I_{LD} = 3A \mbox{ (Note 6)} \\ \hline R_{ISX} = 160\Omega, \\ I_{LD} = 300\mbox{ mA} \end{array}$	-40°C < T <sub>A</sub> < +85°C	-4	+0.9	+4	
			-40°C < T <sub>A</sub> < +125°C	-4	+0.9	+4	]
			-40°C < T <sub>A</sub> < +85°C	-3.5	+0.9	+3.5	1
			-40°C < T <sub>A</sub> < +125°C	-4		+4	1
		R <sub>ISX</sub> = 5.36kΩ,	-40°C < T <sub>A</sub> < +85°C	-3.5	+0.7	+3.5	]
Current Cain Error	6	I <sub>LD</sub> = 30mA	-40°C < T <sub>A</sub> < +125°C	-4		+4	
Current Gain Error	G <sub>I_ERR</sub>	R <sub>ISX</sub> = 160kΩ,	-40°C < T <sub>A</sub> < +85°C	-6	+1.4	+6	- %
		I <sub>LD</sub> = 3mA	-40°C < T <sub>A</sub> < +125°C	-7		+7	
		R <sub>ISX</sub> = 160kΩ,	-40°C < T <sub>A</sub> < +85°C	-12	+1.7	+12	
		I <sub>LD</sub> = 1mA	-40°C < T <sub>A</sub> < +125°C	-15		+15	
		R <sub>ISX</sub> = 160kΩ,	-40°C < T <sub>A</sub> < +85°C	-25	+2.8	+25	
		I <sub>LD</sub> = 300μA	-40°C < T <sub>A</sub> < +125°C	-30		+30	
			R <sub>ISX</sub> = 160Ω, I <sub>LD</sub> = 30mA to 3A		0.4		
Nonlinearity Current Gain Error	G <sub>I_ERR(NON)</sub>	Measured at ISX	R <sub>ISX</sub> = 5.36kΩ, I <sub>LD</sub> = 3mA to 30mA		0.8		%
			R <sub>ISX</sub> = 160kΩ, I <sub>LD</sub> = 300μA to 3mA		1.7		

# 4-Decade Current Sense Amplifier with Internal R<sub>SENSE</sub>

#### **Electrical Characteristics (continued)**

 $(V_{DD} = 3.6V, I_{LD} = 300mA, C_{LD} = 10\mu$ F, SEL0 =  $V_{DD}$ , SEL1 =  $V_{DD}$  (ISH range is selected),  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted. Typical values are at  $T_A = +25^{\circ}$ C (Note 1))

PARAMETER	SYMBOL	CON	NDITIONS	MIN	TYP	MAX	UNITS
		Measured at	R <sub>ISX</sub> = 160Ω, I <sub>LD</sub> = 2A		0.02		
CMRR_ISX (Note 3)		ISX, 0V < V <sub>ISX</sub> <	R <sub>ISX</sub> = 5.36kΩ, I <sub>LD</sub> = 100mA		0.02		%/V
		1.1V	R <sub>ISX</sub> = 160kΩ, I <sub>LD</sub> = 1mA		0.06		
ISX Residual Current		$I_{LD} = 0$			20		nA
AMPLIFIER/DC CHARACTERIS	TICS						
Typical Input Voltage		Guaranteed by C Error	Dutput Amplifier Gain		0.01 to 1.0		V
Offset Voltage	V <sub>OS</sub>	Input referred (N	ote 4)		20		μV
PSRR_VOUT		$\Delta V_{OUT}/\Delta V_{DD}$ , V 2.5V < V <sub>DD</sub> < 5.9	<sub>ISX</sub> = 1.0V, 5V		0.2		mV/V
Output Amplifier Gain	GV				1.5		V/V
Output Amplifier Gain Error	G <sub>V_ERR</sub>	0.01V < V <sub>ISX</sub> < 1	IV	-1	+0.2	+1	%
		$\Delta V_{OUT} / \Delta I_{OUT}$ , s V <sub>ISX</sub> = 1.0V,	ourcing 0 and 2mA,		0.1	1	
Output Load Regulation		$\Delta V_{OUT} / \Delta I_{OUT}$ , s V <sub>ISX</sub> = 10mV	inking 0 and 500µA,		0.1	1	Ω
Leakage Current Into V <sub>OUT</sub> (Low Power Mode)		SEL0 = 0V, SEL	1 = 0V, at V <sub>OUT</sub> = 1.5V		5	100	nA
Max Sink Current		V <sub>ISX</sub> = 0V, V <sub>OUT</sub>	= 1.65V, pulsed test		28		mA
Max Source Current		V <sub>ISX</sub> = 1.1V, V <sub>OI</sub>	<sub>UT</sub> = 0V, pulsed test		28		mA
Total Transimpedance Gain		R <sub>ISX</sub> connected	to ISX pins		0.003 x R <sub>ISX</sub>		
		R <sub>ISX</sub> = 160Ω,	-40°C < T <sub>A</sub> < +85°C	-4	+0.9	+4	
		I <sub>LD</sub> = 3A (Note 6)	-40°C < T <sub>A</sub> < +125°C	-4	+0.9	+4	]
		R <sub>ISX</sub> = 160Ω,	-40°C < T <sub>A</sub> < +85°C	-3.5	1	+3.5	]
		I <sub>LD</sub> = 300mA	-40°C < T <sub>A</sub> < +125°C	-4		+4	
		$R_{ISX} = 5.36 k\Omega$ ,	-40°C < T <sub>A</sub> < +85°C	-3.5	+0.8	+3.5	
Total Transimpedance Gain Error		I <sub>LD</sub> = 30mA	-40°C < T <sub>A</sub> < +125°C	-4		+4	%
(Measured at V <sub>OUT</sub> )		R <sub>ISX</sub> = 160kΩ,	-40°C < T <sub>A</sub> < +85°C	-6	+1.5	+6	
		I <sub>LD</sub> = 3mA	-40°C < T <sub>A</sub> < +125°C	-7		+7	
		R <sub>ISX</sub> = 160kΩ,	-40°C < T <sub>A</sub> < +85°C	-12	+1.8	+12	
		I <sub>LD</sub> = 1mA	-40°C < T <sub>A</sub> < +125°C	-15		+15	
		R <sub>ISX</sub> = 160kΩ,	-40°C < T <sub>A</sub> < +85°C	-25	+3	+25	1
		I <sub>LD</sub> = 300μA	-40°C < T <sub>A</sub> < +125°C	-30		+30	

# 4-Decade Current Sense Amplifier with Internal R<sub>SENSE</sub>

#### **Electrical Characteristics (continued)**

 $(V_{DD} = 3.6V, I_{LD} = 300mA, C_{LD} = 10\mu$ F, SEL0 =  $V_{DD}$ , SEL1 =  $V_{DD}$  (ISH range is selected),  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted. Typical values are at  $T_A = +25$ °C (Note 1))

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
		$R_{ISX} = 160\Omega$ , $I_{LD} = 100$ mA to 3A		0.4		
Nonlinearity Total Transimpedance Gain Error (Measured at V <sub>OUT</sub> )		$R_{ISX} = 5.36 k\Omega$ , $I_{LD} = 3 mA$ to 100mA		0.8		%
		$R_{ISX} = 160 k\Omega$ , $I_{LD} = 300 \mu A$ to 3mA		1.7		
AMPLIFIER/AC CHARACTERIST	TICS	· · · · · · · · · · · · · · · · · · ·				
Concil Cianal Dandwidth		$R_{ISX}$ = 160k $\Omega$ , $I_{LD}$ = 3mA DC and 30 $\mu$ App, $C_{LD}$ = 0		1		
Small Signal Bandwidth		$R_{ISX}$ = 160 $\Omega$ , $I_{LD}$ = 300mA_DC and 3mA_PP, $C_{LD}$ = 0		0.7		- MHz
Large Signal Bandwidth		$R_{ISX}$ = 160Ω, $I_{LD}$ = 2A_DC and 1A_{PP}, $C_{LD}$ = 0		300		kHz
Load Transient Response Time		$R_{ISX}$ = 160kΩ, $I_{LD}$ = 1mA ↔ 2mA		220		μs
		$R_{ISX} = 5.36 k\Omega$ , $I_{LD} = 30 mA \leftrightarrow 60 mA$		70		μs
		$R_{ISX} = 160\Omega, I_{LD} = 1A \leftrightarrow 2A$		60		μs
Output Noise 1/f		0.1Hz to 10Hz		25		μV <sub>PP</sub>
Output Integrated Noise		100Hz to 10kHz		11		μV <sub>RMS</sub>
RANGE SELECT INPUTS (SEL0,	SEL1)					
Input High Level	VIH	SEL0 and SEL1	1			V
Input Low Level	VIL	SEL0 and SEL1			0.5	V
lanut Current	Iн	V <sub>IH</sub> = V <sub>VDD</sub> , SEL0 and SEL1 have weak pulldowns			0.5	
Input Current	IIL	V <sub>IL</sub> = 0V, SEL0 and SEL1 have weak pulldowns			0.5	- μΑ
Low Power Mode, Sleep Delay	t <sub>DIS</sub>	I <sub>LD</sub> = 30mA (Note 5)		5		μs
		R <sub>ISX</sub> = 160Ω, I <sub>LD</sub> = 300mA (Note 5)		30		
Low Power Mode, Waking Delay	t <sub>EN</sub>	R <sub>ISX</sub> = 5.36kΩ, I <sub>LD</sub> = 30mA (Note 5)		50		μs
		$R_{ISX}$ = 160k $\Omega$ , $I_{LD}$ = 1mA (Note 6)		550		
Range Control Delay		Measured from 50% level of SEL0 or SEL1 to the 50% rise of the ISX current		6		μs

Note 1: Limits are 100% tested at  $T_A = +25^{\circ}$ C. Limits over the temperature range and relevant supply voltage range are guaranteed by design and characterization.

Note 2: ISX is any one of the ISL, ISM or ISH pins.

Note 3: CMRR\_ISX is calculated as  $(\Delta I_{ISX} / I_{ISX}) / \Delta V_{ISX}$ .

Note 4: Guaranteed by circuit architecture.

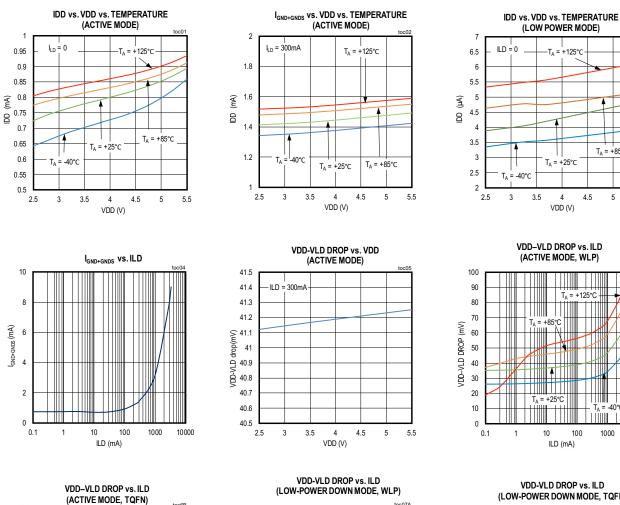
Note 5: Measured from 50% level of SEL0 or SEL1 edge to 50% reduction in the ISX current.

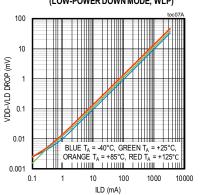
Note 6: Guaranteed by design.

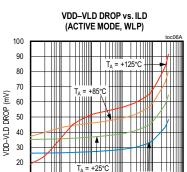
# 4-Decade Current Sense Amplifier with Internal RSENSE

#### **Typical Operating Characteristics**

 $V_{DD} = 3.6V, I_{LD} = 300 \text{mA}, C_{LD} = 10 \mu\text{F}, R_{OUT} = 10 \text{k}\Omega, C_{OUT} = 10 \text{pF}, R_{ISH} = 160\Omega, R_{ISM} = 5.36 \text{k}\Omega, R_{ISL} = 160 \text{k}\Omega \text{ (per the MAX40016 EV kit)}.$ Typical values are at  $T_A = +25^{\circ}C$ , unless otherwise noted.



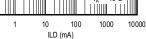




-= +125°0

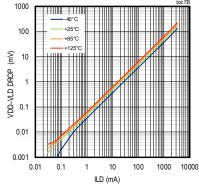
4 4.5 5 5.5

T<sub>A</sub> = +85°C



10.0

VDD-VLD DROP vs. ILD (LOW-POWER DOWN MODE, TQFN)



0.003

0.03

ILD (mA)

0.3

3

+25°C

+85°C

260

240 220

200

180

160

120

100 80

60

40

20

0

0.0003

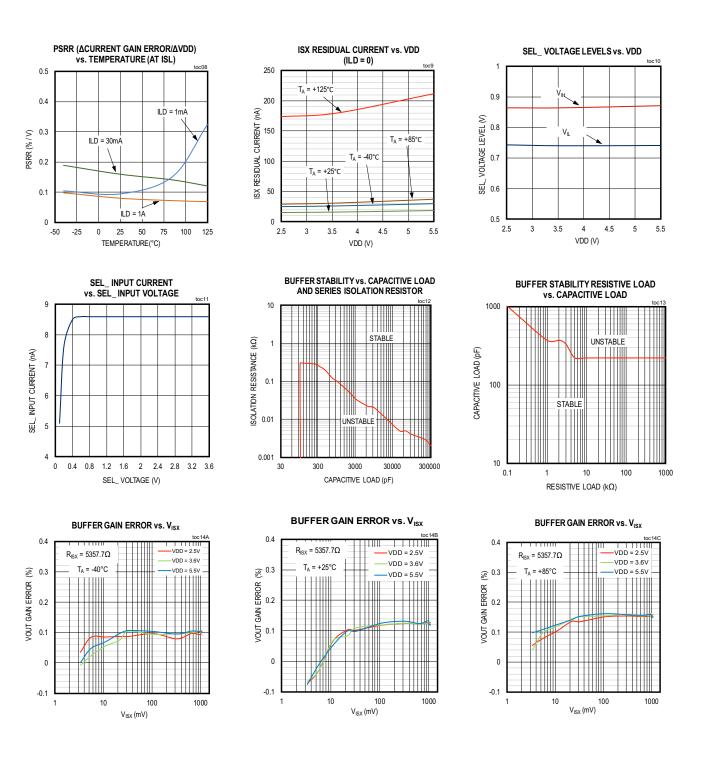
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VDD--VLD DROP 140

# 4-Decade Current Sense Amplifier with Internal R<sub>SENSE</sub>

#### **Typical Operating Characteristics (continued)**

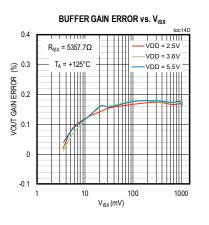
 $V_{DD} = 3.6V, I_{LD} = 300mA, C_{LD} = 10\mu\text{F}, R_{OUT} = 10k\Omega, C_{OUT} = 10p\text{F}, R_{ISH} = 160\Omega, R_{ISM} = 5.36k\Omega, R_{ISL} = 160k\Omega$  (per the MAX40016 EV kit). Typical values are at T<sub>A</sub> = +25°C, unless otherwise noted.

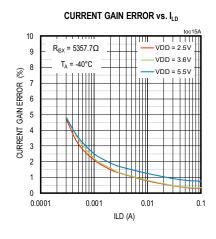


# 4-Decade Current Sense Amplifier with Internal RSENSE

#### **Typical Operating Characteristics (continued)**

 $V_{DD} = 3.6V, I_{LD} = 300 \text{mA}, C_{LD} = 10 \mu\text{F}, R_{OUT} = 10 \text{k}\Omega, C_{OUT} = 10 \text{pF}, R_{ISH} = 160\Omega, R_{ISM} = 5.36 \text{k}\Omega, R_{ISL} = 160 \text{k}\Omega \text{ (per the MAX40016 EV kit)}.$ Typical values are at T<sub>A</sub> = +25°C, unless otherwise noted.





CURRENT GAIN ERROR vs. ILD

5363.7Ω

+125°C

X

0.001

VDD = 2.5V

VDD = 3.6V

VDD = 5.5V

0.1

14

13

12

11

9

8

7 6

5

4

3

2

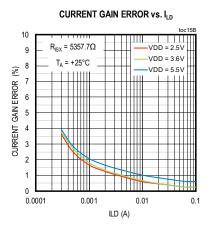
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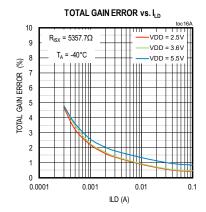
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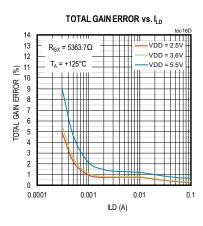
0.0001

(%) 10

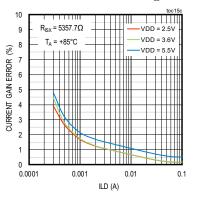
CURRENT GAIN ERROR





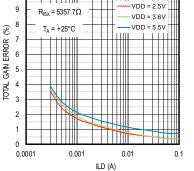


#### CURRENT GAIN ERROR vs. ILD





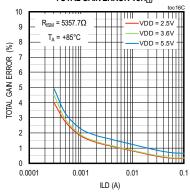
10



TOTAL GAIN ERROR vs. ILD

ILD (A)

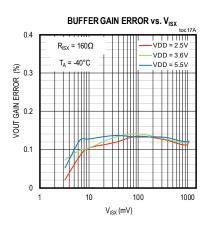
0.01

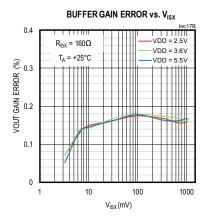


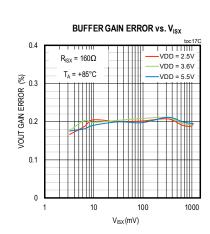
# 4-Decade Current Sense Amplifier with Internal R<sub>SENSE</sub>

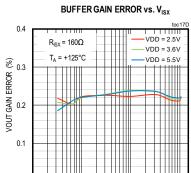
#### **Typical Operating Characteristics (continued)**

 $V_{DD} = 3.6V, I_{LD} = 300mA, C_{LD} = 10\muF, R_{OUT} = 10k\Omega, C_{OUT} = 10pF, R_{ISH} = 160\Omega, R_{ISM} = 5.36k\Omega, R_{ISL} = 160k\Omega$  (per the MAX40016 EV kit). Typical values are at T<sub>A</sub> = +25°C, unless otherwise noted.









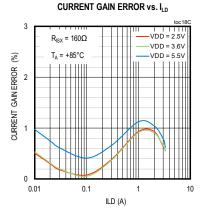
10

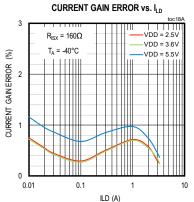
0

1

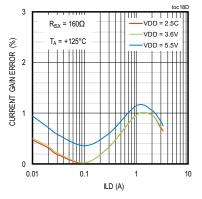
V<sub>ISX</sub> (mV)

100

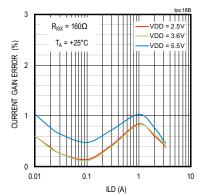




CURRENT GAIN ERROR vs. ILD



CURRENT GAIN ERROR vs. ILD



TOTAL GAIN ERROR vs. ILD 3 R<sub>ISX</sub> = 160Ω VDD = 2.5V VDD = 3.6V T<sub>A</sub> = -40°C VDD = 5.5V TOTAL GAIN ERROR (%) 2 0 0.01 0.1 1 10 ILD (A)

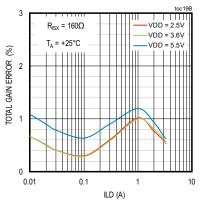
0.01

# 4-Decade Current Sense Amplifier with Internal R<sub>SENSE</sub>

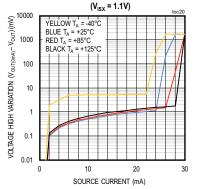
#### **Typical Operating Characteristics (continued)**

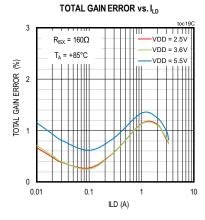
 $V_{DD} = 3.6V, I_{LD} = 300mA, C_{LD} = 10\mu\text{F}, R_{OUT} = 10k\Omega, C_{OUT} = 10p\text{F}, R_{ISH} = 160\Omega, R_{ISM} = 5.36k\Omega, R_{ISL} = 160k\Omega$  (per the MAX40016 EV kit). Typical values are at T<sub>A</sub> = +25°C, unless otherwise noted.

TOTAL GAIN ERROR vs.  $\rm I_{LD}$ 

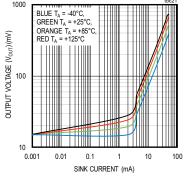


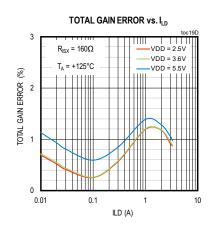
BUFFER OUTPUT VOTLAGE HIGH VARIATION vs. OUTPUT SOURCE CURRENT AT FULL SCALE

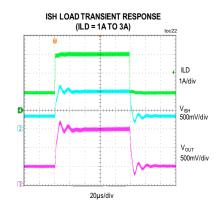


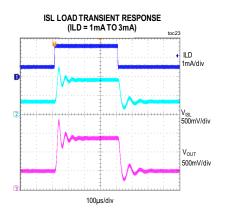


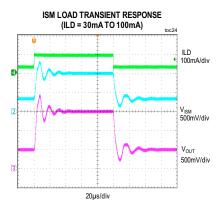
BUFFER OUTPUT VOTLAGE LOW vs. OUTPUT SINK CURRENT (V<sub>Isx</sub> = 10mV)

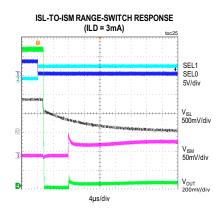








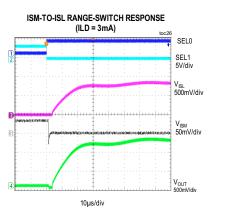


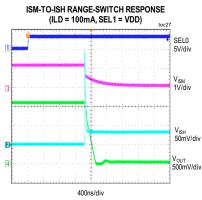


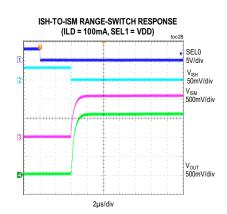
# 4-Decade Current Sense Amplifier with Internal R<sub>SENSE</sub>

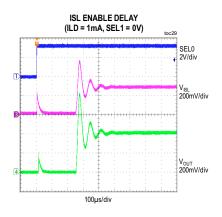
# **Typical Operating Characteristics (continued)**

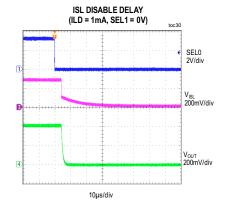
 $V_{DD} = 3.6V, I_{LD} = 300 \text{mA}, C_{LD} = 10 \mu\text{F}, R_{OUT} = 10 \text{k}\Omega, C_{OUT} = 10 \text{pF}, R_{ISH} = 160\Omega, R_{ISM} = 5.36 \text{k}\Omega, R_{ISL} = 160 \text{k}\Omega \text{ (per the MAX40016 EV kit)}.$ Typical values are at T<sub>A</sub> = +25°C, unless otherwise noted.

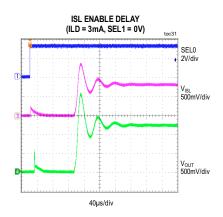


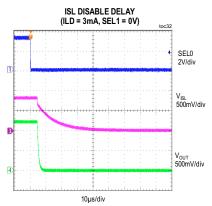


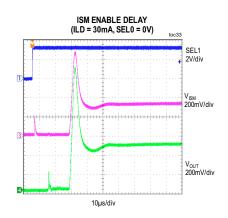


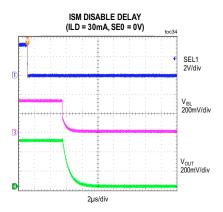








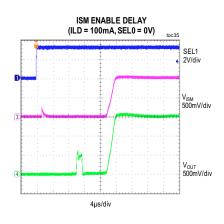


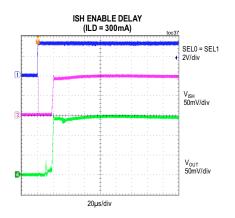


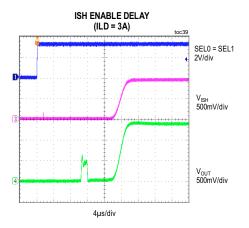
# 4-Decade Current Sense Amplifier with Internal R<sub>SENSE</sub>

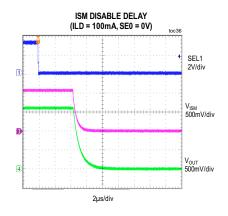
### **Typical Operating Characteristics (continued)**

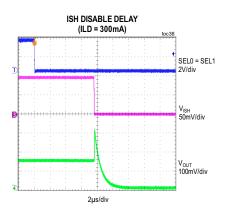
 $V_{DD} = 3.6V, I_{LD} = 300 \text{mA}, C_{LD} = 10 \mu\text{F}, R_{OUT} = 10 \text{k}\Omega, C_{OUT} = 10 \text{pF}, R_{ISH} = 160\Omega, R_{ISM} = 5.36 \text{k}\Omega, R_{ISL} = 160 \text{k}\Omega \text{ (per the MAX40016 EV kit)}.$ Typical values are at T<sub>A</sub> = +25°C, unless otherwise noted.

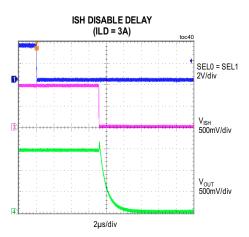










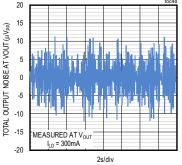


# 4-Decade Current Sense Amplifier with Internal R<sub>SENSE</sub>

#### POWER-DOWN TIME POWER-UP TIME (ILD = 300mA, ISH SELECTED) ISX SMALL-SIGNAL BANDWIDTH (ILD = 300mA, ISH SELECTED) toc42 4 2 VDD VDD 2V/div SMALL-SIGNAL BANDWIDTH (dB) 0 2V/div -2 -4 V<sub>ISH</sub> 50mV/div V<sub>ISH</sub> 50mV/div -6 -8 -10 -12 $I_{LOAD} = 3mA_{DC} \pm 300\mu A_{P-P}$ V<sub>OUT</sub> 100mV/div V<sub>OUT</sub> 100mV/div NORMALIZED TO 2mA/A -14 NO C<sub>LOAD</sub> at LD PIN -16 0.01 0.1 100µs/div 100µs/div BUFFER SMALL-SIGNAL BANDWIDTH ISX SMALL-SIGNAL BANDWIDTH ISX SMALL-SIGNAL BANDWIDTH 10 2 2 150mV<sub>P</sub> 0 0 SMALL-SIGNAL BANDWIDTH (dB) 10kΩ 5 SMALL-SIGNAL BANDWIDTH (dB) SMALL-SIGNAL BANDWIDTH (dB) -2 -2 -4 -4 0 -6 -6 -5 -8 -8 -10 -10 -10 -12 -12 -15 I<sub>LOAD</sub> = 30mA<sub>DC</sub> ±3mA<sub>P-P</sub> NORMALIZED TO 2mA/A $I_{LOAD} = 300 \text{mA}_{DC} \pm 3 \text{mA}_{P-P}$ -14 -14 NORMALIZED TO 2mA/A NO CLOAD at LD PIN NO CLOAD at LD PIN -16 -16 -20 0.01 0.1 0.01 0.1 10 100 1000 10000 0.01 0.1 1 10 100 FREQUENCY (kHz) 1000 10000 1 FREQUENCY (kHz)

BUFFER LARGE-SIGNAL BANDWIDTH 10 Vout RI DAD -= 10kΩ LARGE-SIGNAL BANDWIDTH (dB) 5 0 -5 10 -15 -20 10 100 1000 10000 100000 0.01 0.1 1 FREQUENCY (kHz)

0.1 TO 10 Hz PEAK TO PEAK TOTAL OUTPUT NOISE



TOTAL OUTPUT VOLTAGE NOISE DENSITY vs. FREQUENCY

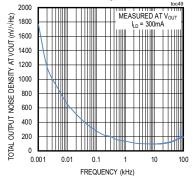
100 10

FREQUENCY (kHz)

1000 10000 100000

10 100 1000 10000

FREQUENCY (kHz)

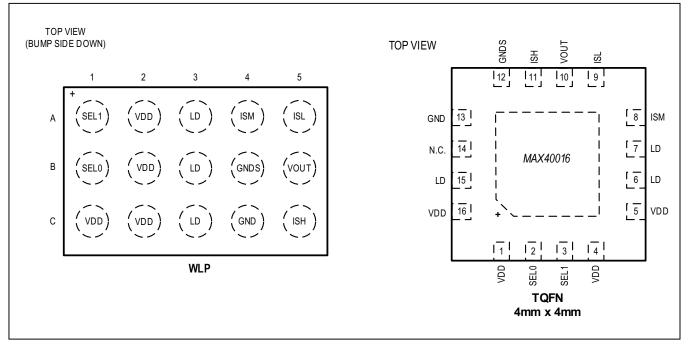


#### **Typical Operating Characteristics (continued)**

 $V_{DD} = 3.6V, I_{LD} = 300 \text{ mA}, C_{LD} = 10 \mu\text{F}, R_{OUT} = 10 k\Omega, C_{OUT} = 10 \text{ pF}, R_{ISH} = 160\Omega, R_{ISM} = 5.36 k\Omega, R_{ISL} = 160 k\Omega$  (per the MAX40016 EV kit). Typical values are at  $T_A = +25^{\circ}C$ , unless otherwise noted.

# 4-Decade Current Sense Amplifier with Internal R<sub>SENSE</sub>

# Pin Configuration



#### **Pin Description**

PI	IN		FUNCTION			
WLP	TQFN	NAME	FUNCTION			
C1, A2, B2, C2	1, 4, 5, 16	V <sub>DD</sub>	Device $V_{DD}$ Supply and Measured Current Input. Bypass $V_{DD}$ to GND with a 0.1µF and a 10µF ceramic capacitors in parallel as close to the device as possible.			
A3, B3, C3	6, 7, 15	LD	Measured Current Output. Connect LD to the load side. Bypass LD to GND with a $10\mu\text{F}$ ceramic capacitor.			
C5	11	ISH*	High Current Range Output. Connect a resistor from ISH to GND to scale the $V_{\mbox{OUT}}$ range.			
A4	8	ISM*	Middle Current Range Output. Connect a resistor from ISM to GND to scale the $V_{OUT}$ range.			
A5	9	ISL*	Low Current Range Output. Connect a resistor from ISL to GND to scale the $V_{\mbox{OUT}}$ range.			
B4	12	GNDS	Ground. Return of the output amplifier's gain setting network. Connect GNDS to GND.			
C4	13	GND	Circuit Ground. All signals are referenced to GND.			
B1	2	SEL0	Logic Selection Input 0 (see Table 1).			
A1	3	SEL1	Logic Selection Input 1 (see Table 1).			
B5	10	V <sub>OUT</sub>	Amplifier Output Voltage. $V_{OUT}$ is proportional to the VDD to ILD current. The scaling factor depends on the resistor values on the ISL, ISM, and ISH inputs.			
—	14	N.C.	No Connect. Internally not connected.			
	EP	EP	Exposed Pad. Internally connected to GND. Connect to a large ground plane to maximize thermal performance. Do not use EP as the only ground connection.			

\*ISL, ISM, and ISH are electrically identical and named to differentiate among the three selectable outputs. Each output when selected is able to support the full-scale sense current range.

# 4-Decade Current Sense Amplifier with Internal R<sub>SENSE</sub>

#### **Detailed Description**

The MAX40016 CSA contains an integrated currentsensing element saving the space and cost of an external sense resistor. Having an integrated sense element has the extra advantage that the entire current measuring path can be factory trimmed, saving the user from having to calibrate independent sense resistors and CSAs.

The CSA has a low power mode in which the currentsensing element remains on, but the output and internal circuitry are turned off to bring the total supply current well below  $10\mu$ A. In this mode, the pass element is turned fully on and will therefore drop slightly less voltage than while it is measuring current. Low power mode is selected by applying a logic-low to both SEL0 and SEL1 (see Table 1).

Three multiplexed scaling outputs from the wide range CSA allow the use of different scaling resistors so that a 12-bit ADC can be sufficient with simple resistor range selection. If only one output is used, an ADC with at least 15 bits of resolution will be needed to realize the full dynamic range of the CSA. See the applications section for details. Each of the scaled outputs are available as a voltage from the V<sub>OUT</sub> pin.

The V<sub>OUT</sub> amplifier output is capable of driving a wide range of ADCs and has a gain of 1.5V/V to provide a full-scale of 1.5V. Most of the values shown in this document are for a full-scale output of 1.5V, suited for 1.8V controllers with embedded 10 to 16-bit ADCs.

The MAX40016 senses from less than  $300\mu$ A to greater than 3A current range. The output maintains less than 5% error specification over a 10,000:1 ratio. In theory, this requires an ADC with a resolution exceeding 13 bits to realize its full dynamic range. While such ADCs are readily available, the system microcontroller already has an embedded 12-bit ADC in many cases.

The three multiplexed scaling current outputs from MAX40016 allow the span to be divided into three ranges that are well within a lower-resolution ADC's capability. Note that it is the same current that is switched to one of the three outputs at a time. The ISH, ISM and ISL pin names are mainly to indicate which output pin is selected. The MAX40016 has its ranges selected using the SEL0 and SEL1 pins. See <u>Current Sense Range Selection</u> (SEL0, SEL1) section and (Table 1) for all the modes.

#### **Scaling Resistors**

The multiplexed scaling resistors' values ( $R_{ISH}$ ,  $R_{ISM}$ ,  $R_{ISL}$ ) should be chosen to suit the ADC's full-scale, usually defined by its reference voltage ( $V_{REF}$ ). Care should be taken to account for all tolerances to avoid overloading the ADC. The typical current from the MAX40016's ISL, or ISM, or ISH pin is specified as 2mA/A. The internal amplifier has a gain of 1.5V/V. Resistors of 0.1% are readily available and so the nominal resistance value is given by:

$$\mathsf{R}_{\mathsf{ISX}} = \frac{\mathsf{VREF}/1.5}{\left(\mathsf{I}_{\mathsf{FS}} \times 0.002\right)} \,(\Omega)$$

The R<sub>ISX</sub> determined from the above equation, where the voltage across the scaling resistor should be limited to 1V, which corresponds to 1.5V full-scale after the amplifier. The closest E192 available value is 167 $\Omega$  which gives very little over-current margin. A 160 $\Omega$  R<sub>ISX</sub> value offers a little more margin towards a conservative design.

#### Current Sense Range Selection (SEL0, SEL1)

SEL0 and SEL1 are digital inputs decoded to control the mirroring of the sense current on the  $V_{DD}$  to LD path to one of three scaled current outputs (ISH, ISM, or ISL), as shown in <u>Table 1</u>. When both SEL0 and SEL1 are at logic 0, the MAX40016 enters its low power operating mode.

SEL0	SEL1	OPERATING MODE/RANGE
0	0	Low Power Mode is Enabled. V <sub>OUT</sub> is high impedance. In low power mode, the current-sensing element still passes current just as an external sense resistor would. There is no capability to turn off the current.
0	1	Middle Current Sense Range (ISM) is Enabled. The resistor R <sub>ISM</sub> connected at this current output termi- nal defines the full-scale voltage of 1V to the internal amplifier.
1	0	Low Current Sense Range (ISL) is Enabled. The resistor R <sub>ISL</sub> connected at this current output terminal defines the full-scale voltage of 1V to the internal amplifier.
1	1	High Current Sense Range (ISH) is Enabled. The resistor R <sub>ISH</sub> connected at this current output terminal defines the full-scale voltage of 1V to the internal amplifier.

#### **Table 1. Current Sense Range Selection**

**Note:** *ISL, ISM, ISH can support all current range from low end to high end. The only difference is that they are selected by different SEL0/SEL1 combination.* 

#### Low Power Mode

The MAX40016 has a low power mode that is activated by pulling both SEL0 and SEL1 low. In this mode, all of the internal circuitry is shut down to save power. The output amplifier is placed in a high impedance state to allow multiplexing of the output line with another MAX40016 for example. In low power mode, the current-sensing element still passes current just as an external sense resistor would. There is no capability to turn off the current.

#### **ISX Residual Current**

When at no load current ( $I_{LD} = 0$ ), there is a small internal residual current at ISX pin due to the internal current mirror block mechanism. This residual current is not an offset current and should not have effect when there is a load current being sensed. Refer to <u>Typical Operating</u> <u>Characteristics</u> for the typical information of this residual current over the the temperature range and V<sub>DD</sub> supply voltage range.

#### **Device Power Up**

Initially, the MAX40016 powers up in low power mode, regardless of the state of SEL0 and SEL1. After the power-up delay time ( $100\mu$ s), the part reverts to the mode selected by SEL0 and SEL1.

#### **Applications Information**

#### **ESD Clamps**

The diagram shows the internal ESD clamping diodes that protect the MAX40016 against electrostatic discharge.

#### **Power Supplies and Bypassing**

The MAX40016 operates from single supply voltage +2.5V to +5.5V. The V<sub>DD</sub> supply input is also the measured current input terminal. Pay extra attention to bypassing and grounding the MAX40016. Peak supply and measured output currents may exceed 3A when the load side experiences large current transients with large external capacitive loads. Supply drops and ground shifts may degrade the

# 4-Decade Current Sense Amplifier with Internal R<sub>SENSE</sub>

device performance. Ground shifts due to insufficient device grounding may also disturb other circuits sharing the same AC ground return path. Any series inductance in the V<sub>DD</sub>, LD and/or GND paths can cause oscillations due to the very high di/dt when switching the MAX40016 with any capacitive load. Bypass V<sub>DD</sub> supply to ground with a 0.1 $\mu$ F in parallel with a 10 $\mu$ F ceramic capacitors as close as possible to the device. Bypass the measured current output, LD terminal, with a 10 $\mu$ F ceramic capacitor or larger depending on the sensing load current, additional bypassing may be needed to keep the device stable during large load output transitions.

#### **Layout Guidelines**

Due to the high currents that may flow through the integrated sensing element based on the application, take care to eliminate solder and parasitic trace resistance from causing errors. Using thicker copper in the PCB construction for these high currents is recommended. Use of Kelvin (force and sense) PCB layout techniques or use of a multilayer PCB with separate ground, power supply and load planes is recommended for noisy digital environments (see the MAX40016EVKIT# data sheet for a layout example). Keep digital signals far away from the sensitive analog inputs. Unshielded long traces at the input and output sense terminals of the device can degrade performance due to noise pick-up.

# 4-Decade Current Sense Amplifier with Internal R<sub>SENSE</sub>

# **Application Information**

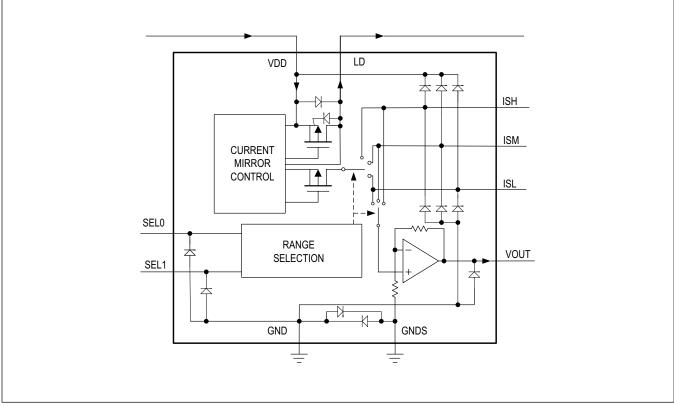


Figure 1. Functional Diagram Showing ESD Clamps

# 4-Decade Current Sense Amplifier with Internal R<sub>SENSE</sub>

#### **Typical Application Circuits**

When the chosen ADC has sufficient resolution to handle the MAX40016 full dynamic range (4-decade of sensing range), only the R<sub>ISH</sub> resistor is required (Figure 2). For a full-scale of 3A the value of R<sub>ISH</sub> is 160 $\Omega$  for a 1V fullscale at the ISH pin, which corresponds to 1.5V output at V<sub>OUT</sub>.

Determining the nominal value of RISH:

The amplifier has a nominal gain of 1.5V/V and the output full-scale voltage is optimized to be 1.5V. So the full-scale voltage across  $R_{\rm ISH}$  is 1V.

The current division factor  $F_{DV}$  (from sensing channel to ISH) is 500 (i.e., 2mA/A).

The full-scale sensed current ( $I_{FS}$ ) is divided by  $F_{DV}$  and the divided current flows through  $R_{ISH}$ .

Thus,  $I_{RSH} = I_{FS}/F_{DV}$ , giving  $R_{ISH} = F_{DV}/I_{FS}$ . Example #1: Using a MAX11214 (24-bit at 64ksps).

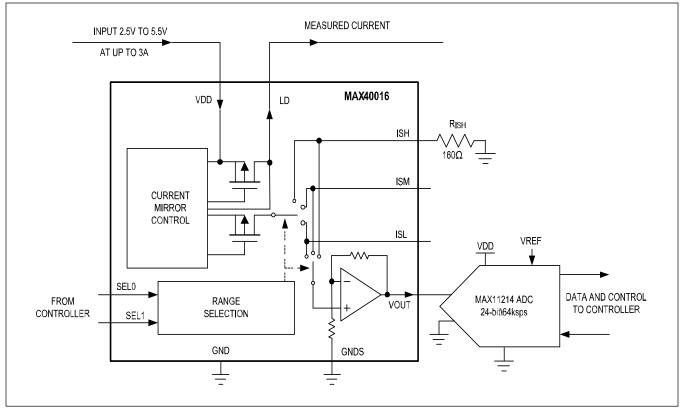


Figure 2. Using the MAX40016 with MAX11214 24-Bit, 64ksps ADC (Single Scaling Resistor with Internal Buffer)

# 4-Decade Current Sense Amplifier with Internal R<sub>SENSE</sub>

The high sampling rate of the MAX11214 renders an anti-aliasing filter unnecessary. Only the R<sub>ISH</sub> resistor is needed to define the gain and the internal programmable gain amplifier inside the ADC allows the selection of reference voltages to match with the 1.5V full-scale from MAX40016. Alternatively, the MAX40016's output buffer can be bypassed and the ADC can be connected directly to the ISH pin, to read the voltage across R<sub>ISH</sub> directly (see Figure 3). If the PCB layout requires a long distance between the MAX40016 and the ADC, the current output from ISH should be run across the PCB and the R<sub>ISH</sub> terminating resistor placed as close as possible to the ADC's input. This helps reduce errors caused by voltage drops across the PCB.

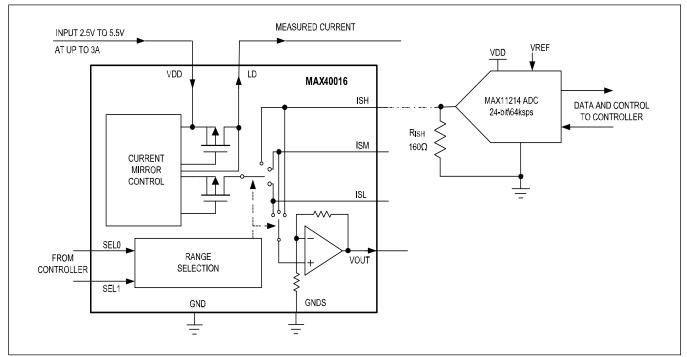


Figure 3. Using the MAX40016 with MAX11214 24-Bit, 64ksps ADC (Single Scaling Resistor without Internal Buffer)

# 4-Decade Current Sense Amplifier with Internal R<sub>SENSE</sub>

#### Implementation with Lower Resolution ADCs

When two or three ranges are required, as in the case of a 10-bit to 12-bit ADC, the higher range resistor (R<sub>ISH</sub>) is calculated as described above. Calculating R<sub>ISM</sub> and or R<sub>ISL</sub> follows the same method with the only difference being the full-scale current is now the lower-range full-scale current. Exactly where it is optimum to arrange this current will depend on the system. Typically splitting the ranges in the region of 30:1 is suitable for most applications. Using R<sub>ISH</sub> = 160 $\Omega$ , R<sub>ISM</sub> = 5.3k $\Omega$ , and R<sub>ISL</sub> = 160k $\Omega$  to split the range up equally (Figure 4). However, this range transition value can be chosen such that the most commonly expected readings would have the better resolution. Selecting too low a transition point leads to more, presumably unnecessary, quantization noise in the higher range.

Example #2: Using an Embedded 12-bit ADC

While typical values for R<sub>IN</sub>, C<sub>IN</sub>, and T<sub>ACQ</sub> are 250 $\Omega$ , 26pF and 350ns these values will vary for different ADCs. Care should be taken to keep all stray capacitance to a minimum.

When the ADC starts its acquisition phase, it suddenly loads the amplifier with  $C_{IN}$  through  $R_{IN}$ . In cases where this loading causes the amplifier's output to ring, an impedance matching filter,  $R_F$  and  $C_F$  may be needed. These components' values should be kept low since they add to the total settling time. Note that Voltage at ISH, ISM, ISL pins should not exceed 1.1V for proper operation (see *Input Voltage Range Under Amplifier* section of the Electrical Characteristics table).

#### **Ordering Information**

PART	TEMP RANGE	PIN- PACKAGE	TOP MARK
MAX40016ANL+T	-40°C to +125°C	15 WLP	+AAB
MAX40016ATE+T	-40°C to +125°C	16 TQFN	—

+Denotes a lead(Pb)-free/RoHS-compliant package. T = Denotes tape-and-reel.

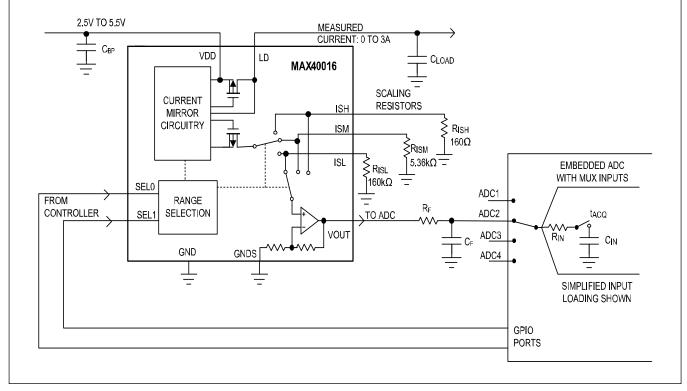


Figure 4. Using the MAX40016 with an Embedded 12-Bit ADC

# 4-Decade Current Sense Amplifier with Internal R<sub>SENSE</sub>

#### **Revision History**

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	1/18	Initial release	—
1	8/18	Updated Benefits and Features, Absolute Maximum Ratings, Package Information, Electrical Characteristics, Typical Operating Characteristics, Pin Description, and Ordering Information	1, 3–5, 7–15, 22

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