

PIC18F87J90 Family Data Sheet

64/80-Pin, High-Performance Microcontrollers with LCD Driver and nanoWatt Technology

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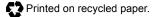
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PIC18F87J90 FAMILY

64/80-Pin, High-Performance Microcontrollers with LCD Driver and nanoWatt Technology

LCD Driver and Keypad Interface Features:

- Direct LCD Panel Drive Capability:
 Can drive LCD panel while in Sleep mode
- Up to 48 Segments and 192 Pixels, Software Selectable
- Programmable LCD Timing module:
 - Multiple LCD timing sources available
- Up to four commons: static, 1/2, 1/3 or 1/4 multiplex
- Static, 1/2 or 1/3 bias configuration
- On-Chip LCD Boost Voltage Regulator for Contrast Control
- Charge Time Measurement Unit (CTMU) for Capacitive Touch Sensing
- ADC for Resistive Touch Sensing

Low-Power Features:

- · Power-Managed modes:
 - Run: CPU On, Peripherals On
 - Idle: CPU Off, Peripherals On
 - Sleep: CPU Off, Peripherals Off
- Two-Speed Oscillator Start-up

Flexible Oscillator Structure:

- Two Crystal modes, 4-25 MHz
- Two External Clock modes, up to 48 MHz
- 4x Phase Lock Loop (PLL)
- Internal Oscillator Block with PLL:
 - Eight user-selectable frequencies from 31.25 kHz to 8 MHz
- Secondary Oscillator using Timer1 at 32 kHz
- Fail-Safe Clock Monitor:
 - Allows for safe shutdown if peripheral clock fails

Peripheral Highlights:

- High-Current Sink/Source 25 mA/25 mA (PORTB and PORTC)
- · Up to Four External Interrupts
- Four 8-Bit/16-Bit Timer/Counter modules
- Two Capture/Compare/PWM (CCP) modules
- Master Synchronous Serial Port (MSSP) module with Two Modes of Operation:
 - 3-Wire/4-Wire SPI (supports all four SPI modes)
- I²C[™] Master and Slave mode
- One Addressable USART module
- One Enhanced Addressable USART module:
 - LIN/J2602 support
 - Auto-wake-up on Start bit and Break character
 - Auto-Baud Detect (ABD)
- 10-Bit, up to 12-Channel A/D Converter:
 - Auto-acquisition
 - Conversion available during Sleep
- Two Analog Comparators
- Programmable Reference Voltage for Comparators
- Hardware Real-Time Clock and Calendar (RTCC) with Clock, Calendar and Alarm Functions
- Charge Time Measurement Unit (CTMU):
 - Capacitance measurement
 - Time measurement with 1 ns typical resolution

Special Microcontroller Features:

- 10,000 Erase/Write Cycle Flash Program Memory, Typical
- Flash Retention 20 Years, Minimum
- Self-Programmable under Software Control
- Word Write Capability for Flash Program Memory for Data EEPROM Emulators

	Flash	SRAM			it »		N	ISSP	RT RT	A/D els)	ators	-VD		
Device	Program Memory (Bytes)	Data Memory (Bytes)	I/O	LCD (Pixels)	Timers 8/16-Bit	ССР	SPI	Master I ² C™	EUSAR AUSAR	10-Bit A (Channe	Compara	BOR/LV	RTCC	CTMU
PIC18F66J90	64K	3,923	51	132	1/3	2	Yes	Yes	1/1	12	2	Yes	Yes	Yes
PIC18F67J90	128K	3,923	51	132	1/3	2	Yes	Yes	1/1	12	2	Yes	Yes	Yes
PIC18F86J90	64K	3,923	67	192	1/3	2	Yes	Yes	1/1	12	2	Yes	Yes	Yes
PIC18F87J90	128K	3,923	67	192	1/3	2	Yes	Yes	1/1	12	2	Yes	Yes	Yes

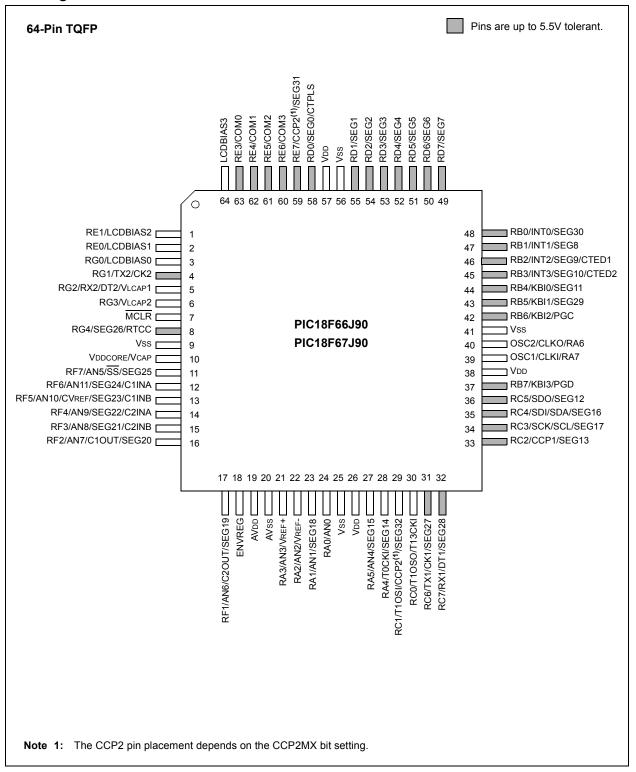
Special Microcontroller Features (Continued):

- · Priority Levels for Interrupts
- 8 x 8 Single-Cycle Hardware Multiplier
- Extended Watchdog Timer (WDT):
 - Programmable period from 4 ms to 131s

- In-Circuit Serial Programming™ (ICSP™) via Two Pins
- In-Circuit Debug via Two Pins
- Operating Voltage Range: 2.0V to 3.6V
- 5.5V Tolerant Input (digital pins only)
- Selectable Open-Drain Configuration for Serial Communication and CCP Pins for Driving Outputs up to 5V
- On-Chip 2.5V Regulator

PIC18F87J90 FAMILY

Pin Diagrams – PIC18F6XJ90



PIC18F87J90 FAMILY

Pin Diagrams – PIC18F8XJ90

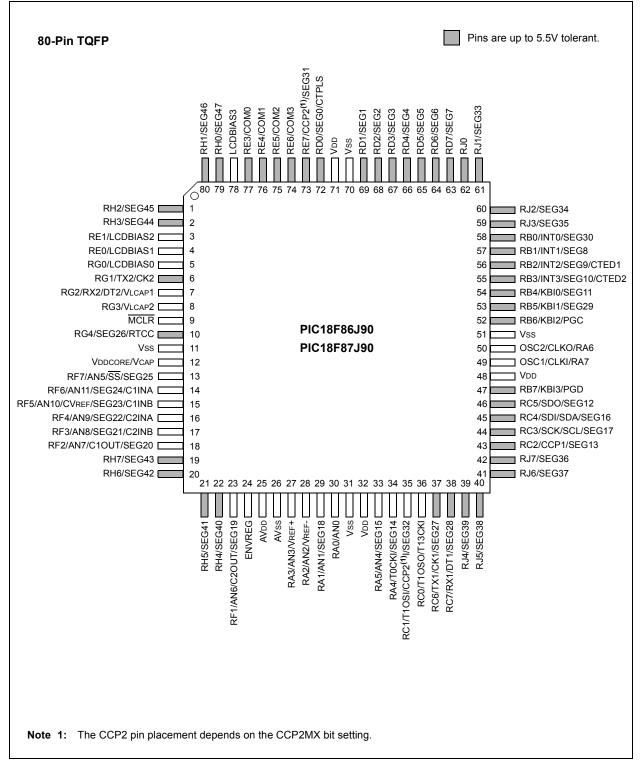


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1.0 DEVICE OVERVIEW

This document contains device-specific information for the following devices:

- PIC18F66J90 PIC18F86J90
- PIC18F67J90 PIC18F87J90

This family combines the traditional advantages of all PIC18 microcontrollers – namely, high computational performance and a rich feature set – with a versatile, on-chip LCD driver, while maintaining an extremely competitive price point. These features make the PIC18F87J90 family a logical choice for many high-performance applications where price is a primary consideration.

1.1 Core Features

1.1.1 nanoWatt TECHNOLOGY

All of the devices in the PIC18F87J90 family incorporate a range of features that can significantly reduce power consumption during operation. Key items include:

- Alternate Run Modes: By clocking the controller from the Timer1 source or the Internal RC oscillator, power consumption during code execution can be reduced by as much as 90%.
- Multiple Idle Modes: The controller can also run with its CPU core disabled but the peripherals still active. In these states, power consumption can be reduced even further, to as little as 4% of normal operation requirements.
- On-the-Fly Mode Switching: The power-managed modes are invoked by user code during operation, allowing the user to incorporate power-saving ideas into their application's software design.

1.1.2 OSCILLATOR OPTIONS AND FEATURES

All of the devices in the PIC18F87J90 family offer six different oscillator options, allowing users a range of choices in developing application hardware. These include:

- Two Crystal modes, using crystals or ceramic resonators.
- Two External Clock modes, offering the option of a divide-by-4 clock output.
- A Phase Lock Loop (PLL) frequency multiplier, available to the External Oscillator modes which allows clock speeds of up to 40 MHz. PLL can also be used with the internal oscillator.
- An internal oscillator block which provides an 8 MHz clock (±2% accuracy) and an INTRC source (approximately 31 kHz, stable over temperature and VDD), as well as a range of six user-selectable clock frequencies, between 125 kHz to 4 MHz, for a total of eight clock frequencies. This option frees the two oscillator pins for use as additional general purpose I/O.

The internal oscillator block provides a stable reference source that gives the family additional features for robust operation:

- Fail-Safe Clock Monitor: This option constantly monitors the main clock source against a reference signal provided by the internal oscillator. If a clock failure occurs, the controller is switched to the internal oscillator, allowing for continued low-speed operation or a safe application shutdown.
- **Two-Speed Start-up:** This option allows the internal oscillator to serve as the clock source from Power-on Reset, or wake-up from Sleep mode, until the primary clock source is available.

1.1.3 MEMORY OPTIONS

The PIC18F87J90 family provides ample room for application code, from 64 Kbytes to 128 Kbytes of code space. The Flash cells for program memory are rated to last up to 10,000 erase/write cycles. Data retention without refresh is conservatively estimated to be greater than 20 years.

The Flash program memory is readable and writable. During normal operation, the PIC18F87J90 family also provides plenty of room for dynamic application data with up to 3,923 bytes of data RAM.

1.1.4 EXTENDED INSTRUCTION SET

The PIC18F87J90 family implements the optional extension to the PIC18 instruction set, adding 8 new instructions and an Indexed Addressing mode. Enabled as a device configuration option, the extension has been specifically designed to optimize re-entrant application code originally developed in high-level languages, such as 'C'.

1.1.5 EASY MIGRATION

Regardless of the memory size, all devices share the same rich set of peripherals, allowing for a smooth migration path as applications grow and evolve.

The consistent pinout scheme used throughout the entire family also aids in migrating to the next larger device. This is true when moving between the 64-pin members, between the 80-pin members, or even jumping from 64-pin to 80-pin devices.

The PIC18F87J90 family is also largely pin-compatible with other PIC18 families, such as the PIC18F8720 and PIC18F8722, the PIC18F85J11, and the PIC18F8490 and PIC18F85J90 families of microcontrollers with LCD drivers. This allows a new dimension to the evolution of applications, allowing developers to select different price points within Microchip's PIC18 portfolio, while maintaining a similar feature set.

1.2 LCD Driver

The on-chip LCD driver includes many features that make the integration of displays in low-power applications easier. These include an integrated voltage regulator with charge pump that allows contrast control in software and display operation above device VDD.

1.3 Other Special Features

- Communications: The PIC18F87J90 family incorporates a range of serial communication peripherals, including an Addressable USART, a separate Enhanced USART that supports LIN/J2602 specification 1.2, and one Master SSP module capable of both SPI and I²C[™] (Master and Slave) modes of operation.
- CCP Modules: All devices in the family incorporate two Capture/Compare/PWM (CCP) modules. Up to four different time bases may be used to perform several different operations at once.
- **10-Bit A/D Converter:** This module incorporates programmable acquisition time, allowing for a channel to be selected and a conversion to be initiated without waiting for a sampling period and thus, reducing code overhead.
- Charge Time Measurement Unit (CTMU): The CTMU is a flexible analog module that provides accurate differential time measurement between pulse sources, as well as asynchronous pulse generation.

Together with other on-chip analog modules, the CTMU can precisely measure time, measure capacitance or relative changes in capacitance, or generate output pulses that are independent of the system clock.

- Extended Watchdog Timer (WDT): This enhanced version incorporates a 16-bit prescaler, allowing an extended time-out range that is stable across operating voltage and temperature. See Section 28.0 "Electrical Characteristics" for time-out periods.
- Real-Time Clock and Calendar Module (RTCC): The RTCC module is intended for applications requiring that accurate time be maintained for extended periods of time with minimum to no intervention from the CPU.

The module is a 100-year clock and calendar with automatic leap year detection. The range of the clock is from 00:00:00 (midnight) on January 1, 2000 to 23:59:59 on December 31, 2099.

1.4 Details on Individual Family Members

Devices in the PIC18F87J90 family are available in 64-pin and 80-pin packages. Block diagrams for the two groups are shown in Figure 1-1 and Figure 1-2.

The devices are differentiated from each other in four ways:

- Flash program memory (two sizes, 64 Kbytes for PIC18FX6J90 devices and 128 Kbytes for PIC18FX7J90 devices).
- 2. Data RAM (3,923 bytes RAM for both PIC18FX6J90 and PIC18FX7J90 devices).
- I/O ports (7 bidirectional ports on PIC18F6XJ90 devices, 9 bidirectional ports on PIC18F8XJ90 devices).
- LCD Pixels: 132 pixels (33 SEGs x 4 COMs) can be driven by 64-pin devices; 192 pixels (48 SEGs x 4 COMs) can be driven by 80-pin devices.

All other features for devices in this family are identical. These are summarized in Table 1-1 and Table 1-2.

The pinouts for all devices are listed in Table 1-3 and Table 1-4.

Features	PIC18F66J90	PIC18F67J90
Operating Frequency	DC - 4	8 MHz
Program Memory (Bytes)	64K	128K
Program Memory (Instructions)	32,768	65,536
Data Memory (Bytes)	3,923	3,923
Interrupt Sources	29	9
I/O Ports	Ports A, B, C	C, D, E, F, G
LCD Driver (available pixels to drive)	132 (33 SEG	s x 4 COMs)
Timers	4	
Comparators	2	2
СТМИ	Ye	es
RTCC	Ye	es
Capture/Compare/PWM Modules	2	2
Serial Communications	MSSP, Addressable USA	ART, Enhanced USART
10-Bit Analog-to-Digital Module	12 Input C	Channels
Resets (and Delays)	POR, BOR, RESET Instruction, Stack (PWRT	
Instruction Set	75 Instructions, 83 with Exter	nded Instruction Set Enabled
Packages	64-Pin	TQFP

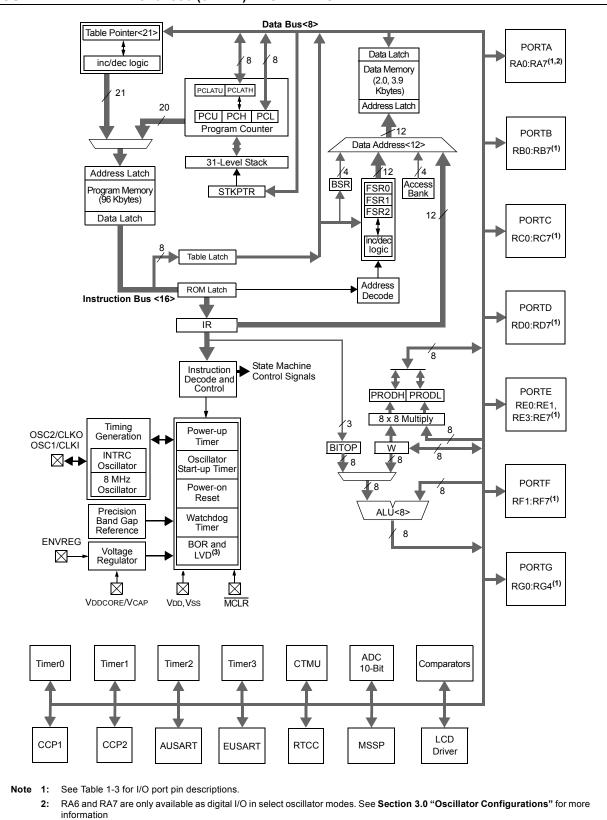
TABLE 1-1: DEVICE FEATURES FOR THE PIC18F6XJ90 (64-PIN DEVICES)

TABLE 1-2: DEVICE FEATURES FOR THE PIC18F8XJ90 (80-PIN DEVICES)

Features	PIC18F86J90	PIC18F87J90
Operating Frequency	DC - 4	8 MHz
Program Memory (Bytes)	64K	128K
Program Memory (Instructions)	32,768	65,536
Data Memory (Bytes)	3,923	3,923
Interrupt Sources	2	9
I/O Ports	Ports A, B, C, I	D, E, F, G, H, J
LCD Driver (available pixels to drive)	192 (48 SEG	s x 4 COMs)
Timers	4	
Comparators	2	2
СТМИ	Ye	es
RTCC	Ye	es
Capture/Compare/PWM Modules	2	2
Serial Communications	MSSP, Addressable US	ART, Enhanced USART
10-Bit Analog-to-Digital Module	12 Input 0	Channels
Resets (and Delays)	POR, BOR, RESET Instruction, Stack (PWR)	
Instruction Set	75 Instructions, 83 with Exter	nded Instruction Set Enabled
Packages	80-Pin	TQFP

PIC18F87J90 FAMILY





3: Brown-out Reset and Low-Voltage Detect functions are provided when the on-board voltage regulator is enabled.

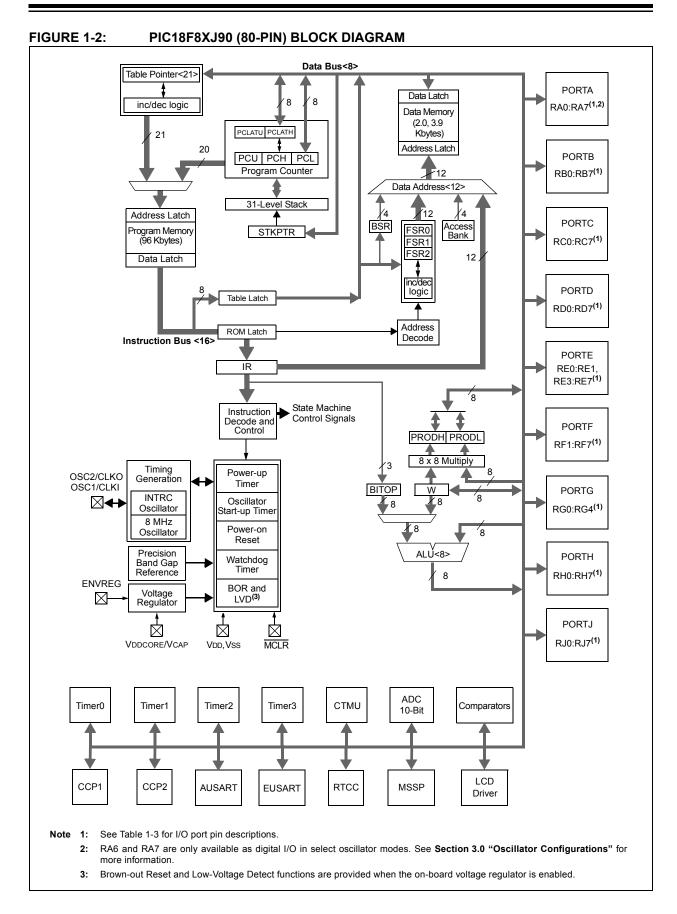


TABLE 1-3: PIC18F6XJ90 PINOUT I/O DESCRIPTIONS

Pin Name	Pin Number	Pin	Buffer	Description
	TQFP	Туре	Туре	Description
MCLR	7	I	ST	Master Clear (input) or programming voltage (input). This pin is an active-low Reset to the device.
OSC1/CLKI/RA7 OSC1 CLKI RA7	39	 /0	CMOS CMOS TTL	Oscillator crystal or external clock input. Oscillator crystal input. External clock source input. Always associated with pin function, OSC1. (See related OSC1/CLKI, OSC2/CLKO pins.) General purpose I/O pin.
OSC2/CLKO/RA6 OSC2	40	ο	_	Oscillator crystal or clock output. Oscillator crystal output. Connects to crystal or resonator in Crystal Oscillator mode.
CLKO		0	_	In EC modes, OSC2 pin outputs CLKO, which has 1/4 the frequency of OSC1 and denotes the instruction cycle rate.
RA6		I/O	TTL	General purpose I/O pin.
				PORTA is a bidirectional I/O port.
RA0/AN0 RA0 AN0	24	I/O I	TTL Analog	Digital I/O. Analog Input 0.
RA1/AN1/SEG18 RA1 AN1 SEG18	23	I/O I O	TTL Analog Analog	Digital I/O. Analog Input 1. SEG18 output for LCD.
RA2/AN2/VREF- RA2 AN2 VREF-	22	I/O I I	TTL Analog Analog	Digital I/O. Analog Input 2. A/D reference voltage (low) input.
RA3/AN3/Vref+ RA3 AN3 Vref+	21	I/O I I	TTL Analog Analog	Digital I/O. Analog Input 3. A/D reference voltage (high) input.
RA4/T0CKI/SEG14 RA4 T0CKI SEG14	28	I/O I O	ST ST Analog	Digital I/O. Timer0 external clock input. SEG14 output for LCD.
RA5/AN4/SEG15 RA5 AN4 SEG15	27	I/O I O	TTL Analog Analog	Digital I/O. Analog Input 4. SEG15 output for LCD.
RA6				See the OSC2/CLKO/RA6 pin.
RA7				See the OSC1/CLKI/RA7 pin.
Legend: TTL = TTL c ST = Schm I = Input P = Powe $I^2C^{TM} = I^2C/S$	ompatible input itt Trigger input r MBus	with C	MOS leve	CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

Note 1: Default assignment for CCP2 when the CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin	Buffer	Description
Fill Name	TQFP	Туре	Туре	Description
				PORTB is a bidirectional I/O port. PORTB can be software programmed for internal weak pull-ups on all inputs.
RB0/INT0/SEG30 RB0 INT0 SEG30	48	I/O I O	TTL ST Analog	Digital I/O. External Interrupt 0. SEG30 output for LCD.
RB1/INT1/SEG8 RB1 INT1 SEG8	47	I/O I O	TTL ST Analog	Digital I/O. External Interrupt 1. SEG8 output for LCD.
RB2/INT2/SEG9/CTED1 RB2 INT2 CTED1 SEG9	46	I/O I I O	TTL ST ST Analog	Digital I/O. External Interrupt 2. CTMU Edge 1 input. SEG9 output for LCD.
RB3/INT3/SEG10/CTED2 RB3 INT3 SEG10 CTED2	45	I/O I O I	TTL ST Analog ST	Digital I/O. External Interrupt 3. SEG10 output for LCD. CTMU Edge 2 input.
RB4/KBI0/SEG11 RB4 KBI0 SEG11	44	I/O I O	TTL TTL Analog	Digital I/O. Interrupt-on-change pin. SEG11 output for LCD.
RB5/KBI1/SEG29 RB5 KBI1 SEG29	43	I/O I O	TTL TTL Analog	Digital I/O. Interrupt-on-change pin. SEG29 output for LCD.
RB6/KBI2/PGC RB6 KBI2 PGC	42	I/O I I/O	TTL TTL ST	Digital I/O. Interrupt-on-change pin. In-Circuit Debugger and ICSP™ programming clock pin
RB7/KBI3/PGD RB7 KBI3 PGD	37	I/O I I/O	TTL TTL ST	Digital I/O. Interrupt-on-change pin. In-Circuit Debugger and ICSP programming data pin.
I = Input P = Power $I^2C^{TM} = I^2C/SN$	t Trigger input v IBus			CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

TABLE 1-3: PIC18F6XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)

Note 1: Default assignment for CCP2 when the CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin	Buffer	Description
FIII Ndille	TQFP	Туре	Туре	Description
				PORTC is a bidirectional I/O port.
RC0/T1OSO/T13CKI RC0 T1OSO T13CKI	30	I/O O I	ST — ST	Digital I/O. Timer1 oscillator output. Timer1/Timer3 external clock input.
RC1/T1OSI/CCP2/SEG32 RC1 T1OSI CCP2 ⁽¹⁾ SEG32	29	I/O I I/O O	ST CMOS ST Analog	Digital I/O. Timer1 oscillator input. Capture 2 input/Compare 2 output/PWM2 output. SEG32 output for LCD.
RC2/CCP1/SEG13 RC2 CCP1 SEG13	33	I/O I/O O	ST ST Analog	Digital I/O. Capture 1 input/Compare 1 output/PWM1 output. SEG13 output for LCD.
RC3/SCK/SCL/SEG17 RC3 SCK SCL SEG17	34	I/O I/O I/O O	ST ST I ² C Analog	Digital I/O. Synchronous serial clock input/output for SPI mode. Synchronous serial clock input/output for I ² C™ mode. SEG17 output for LCD.
RC4/SDI/SDA/SEG16 RC4 SDI SDA SEG16	35	I/O I I/O O	ST ST I ² C Analog	Digital I/O. SPI data in. I ² C data I/O. SEG16 output for LCD.
RC5/SDO/SEG12 RC5 SDO SEG12	36	I/O O O	ST Analog	Digital I/O. SPI data out. SEG12 output for LCD.
RC6/TX1/CK1/SEG27 RC6 TX1 CK1 SEG27	31	I/O O I/O O	ST — ST Analog	Digital I/O. EUSART asynchronous transmit. EUSART synchronous clock (see related RX1/DT1). SEG27 output for LCD.
RC7/RX1/DT1/SEG28 RC7 RX1 DT1 SEG28	32	I/O I I/O O	ST ST ST Analog	Digital I/O. EUSART asynchronous receive. EUSART synchronous data (see related TX1/CK1). SEG28 output for LCD.
Legend: TTL = TTL co ST = Schmit I = Input P = Power $I^2C^{TM} = I^2C/SM$	t Trigger input	with C	MOS leve	CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

TABLE 1-3: PIC18F6XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)

Note 1: Default assignment for CCP2 when the CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin	Buffer	Description
Pin Name	TQFP	Туре	Туре	Description
				PORTD is a bidirectional I/O port.
RD0/SEG0/CTPLS RD0 SEG0 CTPLS	58	I/O O O	ST Analog —	Digital I/O. SEG0 output for LCD. CTMU pulse generator output.
RD1/SEG1 RD1 SEG1	55	I/O O	ST Analog	Digital I/O. SEG1 output for LCD.
RD2/SEG2 RD2 SEG2	54	I/O O	ST Analog	Digital I/O. SEG2 output for LCD.
RD3/SEG3 RD3 SEG3	53	I/O O	ST Analog	Digital I/O. SEG3 output for LCD.
RD4/SEG4 RD4 SEG4	52	I/O O	ST Analog	Digital I/O. SEG4 output for LCD.
RD5/SEG5 RD5 SEG5	51	I/O O	ST Analog	Digital I/O. SEG5 output for LCD.
RD6/SEG6 RD6 SEG6	50	I/O O	ST Analog	Digital I/O. SEG6 output for LCD.
RD7/SEG7 RD7 SEG7	49	I/O O	ST Analog	Digital I/O. SEG7 output for LCD.
Legend: TTL = TTL co ST = Schmit I = Input P = Power I^2C^{TM} = I^2C/SM	t Trigger input v	with Cl	MOS level	CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

TABLE 1-3:	PIC18F6XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)

Note 1: Default assignment for CCP2 when the CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin	Buffer	Description
Pin Name	TQFP	Туре	Туре	Description
				PORTE is a bidirectional I/O port.
RE0/LCDBIAS1 RE0 LCDBIAS1	2	I/O I	ST Analog	Digital I/O. BIAS1 input for LCD.
RE1/LCDBIAS2 RE1 LCDBIAS2	1	I/O I	ST Analog	Digital I/O. BIAS2 input for LCD.
LCDBIAS3	64	Т	Analog	BIAS3 input for LCD.
RE3/COM0 RE3 COM0	63	I/O O	ST Analog	Digital I/O. COM0 output for LCD.
RE4/COM1 RE4 COM1	62	I/O O	ST Analog	Digital I/O. COM1 output for LCD.
RE5/COM2 RE5 COM2	61	I/O O	ST Analog	Digital I/O. COM2 output for LCD.
RE6/COM3 RE6 COM3	60	I/O O	ST Analog	Digital I/O. COM3 output for LCD.
RE7/CCP2/SEG31 RE7 CCP2 ⁽²⁾ SEG31	59	I/O I/O O	ST ST Analog	Digital I/O. Capture 2 input/Compare 2 output/PWM2 output. SEG31 output for LCD.
Legend: TTL = TTL co ST = Schmi I = Input P = Power $I^2C^{TM} = I^2C/SM$	tt Trigger input v MBus			CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

TABLE 1-3: PIC18F6XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)

Note 1: Default assignment for CCP2 when the CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin	Buffer	Description
Fill Name	TQFP	Туре	Туре	Description
				PORTF is a bidirectional I/O port.
RF1/AN6/C2OUT/SEG19 RF1 AN6 C2OUT SEG19	17	I/O I O O	ST Analog — Analog	Digital I/O. Analog Input 6. Comparator 2 output. SEG19 output for LCD.
RF2/AN7/C1OUT/SEG20 RF2 AN7 C1OUT SEG20	16	I/O I O O	ST Analog — Analog	Digital I/O. Analog Input 7. Comparator 1 output. SEG20 output for LCD.
RF3/AN8/SEG21/C2INB RF3 AN8 SEG21 C2INB	15	I/O I O I	ST Analog Analog Analog	Digital I/O. Analog Input 8. SEG21 output for LCD. Comparator 2 Input B.
RF4/AN9/SEG22/C2INA RF4 AN9 SEG22 C2INA	14	I/O I O I	ST Analog Analog Analog	Digital I/O. Analog Input 9. SEG22 output for LCD Comparator 2 Input A.
RF5/AN10/CVREF/ SEG23/C1INB RF5 AN10 CVREF SEG23 C1INB	13	I/O I O I	ST Analog Analog Analog Analog	Digital I/O. Analog Input 10. Comparator reference voltage output. SEG23 output for LCD. Comparator 1 Input B.
RF6/AN11/SEG24/C1INA RF6 AN11 SEG24 C1INA	12	I/O I O I	ST Analog Analog Analog	Digital I/O. Analog Input 11. SEG24 output for LCD Comparator 1 Input A.
RF7/AN5/SS/SEG25 RF7 AN5 SS SEG25	11	I/O O I O	ST Analog TTL Analog	Digital I/O. Analog Input 5. SPI slave select input. SEG25 output for LCD.
Legend: TTL = TTL co ST = Schmit I = Input P = Power	t Trigger input v	with C	MOS leve	CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

TABLE 1-3: PIC18F6XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)
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 $I^2C^{TM} = I^2C/SMBus$ Note 1: Default assignment for CCP2 when the CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin	Buffer Type	Description
Pin Name	TQFP	Туре		
				PORTG is a bidirectional I/O port.
RG0/LCDBIAS0 RG0 LCDBIAS0	3	I/O I	ST Analog	Digital I/O. BIAS0 input for LCD.
RG1/TX2/CK2 RG1 TX2 CK2	4	I/O O I/O	ST — ST	Digital I/O. AUSART asynchronous transmit. AUSART synchronous clock (see related RX2/DT2).
RG2/RX2/DT2/VLCAP1 RG2 RX2 DT2 VLCAP1	5	I/O I I/O I	ST ST ST Analog	Digital I/O. AUSART asynchronous receive. AUSART synchronous data (see related TX2/CK2). LCD charge pump capacitor input.
RG3/VLCAP2 RG3 VLCAP2	6	I/O I	ST Analog	Digital I/O. LCD charge pump capacitor input.
RG4/SEG26/RTCC RG4 SEG26 RTCC	8	I/O O O	ST Analog —	Digital I/O. SEG26 output for LCD. RTCC output
Vss	9, 25, 41, 56	Р	_	Ground reference for logic and I/O pins.
Vdd	26, 38, 57	Р		Positive supply for logic and I/O pins.
AVss	20	Р		Ground reference for analog modules.
AVdd	19	Р		Positive supply for analog modules.
ENVREG	18	Ι	ST	Enable for on-chip voltage regulator.
VDDCORE/VCAP VDDCORE	10	Р	_	Core logic power or external filter capacitor connection. Positive supply for microcontroller core logic (regulator disabled).
VCAP		Р	—	External filter capacitor connection (regulator enabled).
Legend: TTL = TTL cc ST = Schmi I = Input P = Power $I^2C^{TM} = I^2C/SN$	tt Trigger input	with C	MOS leve	CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

TABLE 1-3: PIC18F6XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)

Note 1: Default assignment for CCP2 when the CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin		Description
FIII Naille	TQFP	Туре		
MCLR	9	I	ST	Master Clear (input) or programming voltage (input). This pin is an active-low Reset to the device.
OSC1/CLKI/RA7 OSC1 CLKI RA7	49	 /0	CMOS CMOS TTL	Oscillator crystal or external clock input. Oscillator crystal input. External clock source input. Always associated with pin function, OSC1. (See related OSC1/CLKI, OSC2/CLKO pins.) General purpose I/O pin.
OSC2/CLKO/RA6	50	1/0	116	Oscillator crystal or clock output.
OSC2	50	0	—	Oscillator crystal output. Connects to crystal or resonator in Crystal Oscillator mode.
CLKO		0	_	In EC modes, OSC2 pin outputs CLKO, which has 1/4 the frequency of OSC1 and denotes the instruction cycle rate.
RA6		I/O	TTL	General purpose I/O pin.
				PORTA is a bidirectional I/O port.
RA0/AN0 RA0 AN0	30	I/O I	TTL Analog	Digital I/O. Analog Input 0.
RA1/AN1/SEG18 RA1 AN1 SEG18	29	I/O I O	TTL Analog Analog	Digital I/O. Analog Input 1. SEG18 output for LCD.
RA2/AN2/VREF- RA2 AN2 VREF-	28	I/O I I	TTL Analog Analog	Digital I/O. Analog Input 2. A/D reference voltage (low) input.
RA3/AN3/Vref+ RA3 AN3 Vref+	27	I/O I I	TTL Analog Analog	Digital I/O. Analog Input 3. A/D reference voltage (high) input.
RA4/T0CKI/SEG14 RA4 T0CKI SEG14	34	I/O I O	ST ST Analog	Digital I/O. Timer0 external clock input. SEG14 output for LCD.
RA5/AN4/SEG15 RA5 AN4 SEG15	33	I/O I O	TTL Analog Analog	Digital I/O. Analog Input 4. SEG15 output for LCD.
RA6				See the OSC2/CLKO/RA6 pin.
RA7				See the OSC1/CLKI/RA7 pin.
	er	with C	MOS leve	CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

TABLE 1-4:	PIC18F8XJ90 PINOUT I/O DESCRIPTIONS

Note 1: Default assignment for CCP2 when the CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin	Buffer	Description
Pin Name	TQFP	Туре	Туре Туре	
				PORTB is a bidirectional I/O port. PORTB can be software programmed for internal weak pull-ups on all inputs.
RB0/INT0/SEG30 RB0 INT0 SEG30	58	I/O I O	TTL ST Analog	Digital I/O. External Interrupt 0. SEG30 output for LCD.
RB1/INT1/SEG8 RB1 INT1 SEG8	57	I/O I O	TTL ST Analog	Digital I/O. External Interrupt 1. SEG8 output for LCD.
RB2/INT2/SEG9/CTED1 RB2 INT2 SEG9 CTED1	56	I/O I O I	TTL ST Analog ST	Digital I/O. External Interrupt 2. SEG9 output for LCD. CTMU Edge 1 input.
RB3/INT3/SEG10/ CTED2 RB3 INT3 SEG10 CTED2	55	I/O I O I	TTL ST Analog ST	Digital I/O. External Interrupt 3. SEG10 output for LCD. CTMU Edge 2 input.
RB4/KBI0/SEG11 RB4 KBI0 SEG11	54	I/O I O	TTL TTL Analog	Digital I/O. Interrupt-on-change pin. SEG11 output for LCD.
RB5/KBI1/SEG29 RB5 KBI1 SEG29	53	I/O I O	TTL TTL Analog	Digital I/O. Interrupt-on-change pin. SEG29 output for LCD.
RB6/KBI2/PGC RB6 KBI2 PGC	52	I/O I I/O	TTL TTL ST	Digital I/O. Interrupt-on-change pin. In-Circuit Debugger and ICSP™ programming clock pin.
RB7/KBI3/PGD RB7 KBI3 PGD	47	I/O I I/O	TTL TTL ST	Digital I/O. Interrupt-on-change pin. In-Circuit Debugger and ICSP programming data pin.
		with C	MOS leve	CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

TABLE 1-4: PIC18F8XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)

P = Power I²C™ = I²C/SMBus

Note 1: Default assignment for CCP2 when the CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin	Buffer	Description
Fill Name	TQFP	Туре Туре		Description
				PORTC is a bidirectional I/O port.
RC0/T10SO/T13CKI RC0 T10SO T13CKI	36	I/O O I	ST — ST	Digital I/O. Timer1 oscillator output. Timer1/Timer3 external clock input.
RC1/T1OSI/CCP2/SEG32 RC1 T1OSI CCP2 ⁽¹⁾ SEG32	35	I/O I I/O O	ST CMOS ST Analog	Digital I/O. Timer1 oscillator input. Capture 2 input/Compare 2 output/PWM2 output. SEG32 output for LCD.
RC2/CCP1/SEG13 RC2 CCP1 SEG13	43	I/O I/O O	ST ST Analog	Digital I/O. Capture 1 input/Compare 1 output/PWM1 output. SEG13 output for LCD.
RC3/SCK/SCL/SEG17 RC3 SCK SCL SEG17	44	I/O I/O I/O O	ST ST I ² C Analog	Digital I/O. Synchronous serial clock input/output for SPI mode. Synchronous serial clock input/output for I ² C™ mode. SEG17 output for LCD.
RC4/SDI/SDA/SEG16 RC4 SDI SDA SEG16	45	I/O I I/O O	ST ST I ² C Analog	Digital I/O. SPI data in. I ² C data I/O. SEG16 output for LCD.
RC5/SDO/SEG12 RC5 SDO SEG12	46	I/O O O	ST — Analog	Digital I/O. SPI data out. SEG12 output for LCD.
RC6/TX1/CK1/SEG27 RC6 TX1 CK1 SEG27	37	I/O O I/O O	ST — ST Analog	Digital I/O. EUSART asynchronous transmit. EUSART synchronous clock (see related RX1/DT1). SEG27 output for LCD.
RC7/RX1/DT1/SEG28 RC7 RX1 DT1 SEG28	38	1/0 1/0 0	ST ST ST Analog	Digital I/O. EUSART asynchronous receive. EUSART synchronous data (see related TX1/CK1). SEG28 output for LCD.
Legend: TTL = TTL cc ST = Schmit I = Input P = Power I^2C^{TM} = I ² C/SM	t Trigger input	with C	MOS leve	CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

TABLE 1-4: PIC18F8XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)

Note 1: Default assignment for CCP2 when the CCP2MX Configuration bit is set.

Din Nome	Pin Number	Pin		
Pin Name	TQFP	Туре		Description
				PORTD is a bidirectional I/O port.
RD0/SEG0/CTPLS RD0 SEG0 CTPLS	72	I/O O O	ST Analog ST	Digital I/O. SEG0 output for LCD. CTMU pulse generator output.
RD1/SEG1 RD1 SEG1	69	I/O O	ST Analog	Digital I/O. SEG1 output for LCD.
RD2/SEG2 RD2 SEG2	68	I/O O	ST Analog	Digital I/O. SEG2 output for LCD.
RD3/SEG3 RD3 SEG3	67	I/O O	ST Analog	Digital I/O. SEG3 output for LCD.
RD4/SEG4 RD4 SEG4	66	I/O O	ST Analog	Digital I/O. SEG4 output for LCD.
RD5/SEG5 RD5 SEG5	65	I/O O	ST Analog	Digital I/O. SEG5 output for LCD.
RD6/SEG6 RD6 SEG6	64	I/O O	ST Analog	Digital I/O. SEG6 output for LCD.
RD7/SEG7 RD7 SEG7	63	I/O O	ST Analog	Digital I/O. SEG7 output for LCD.
Legend: TTL = TTL cc ST = Schmi I = Input P = Power	tt Trigger input	with C	MOS leve	CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

TABLE 1-4: PIC18F8XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)

Note 1: Default assignment for CCP2 when the CCP2MX Configuration bit is set.

 $I^2C^{TM} = I^2C/SMBus$

	Pin Number	Pin	Buffer	Description
Pin Name	TQFP	Туре Туре		Description
				PORTE is a bidirectional I/O port.
RE0/LCDBIAS1 RE0 LCDBIAS1	4	I/O I	ST Analog	Digital I/O. BIAS1 input for LCD.
RE1/LCDBIAS2 RE1 LCDBIAS2	3	I/O I	ST Analog	Digital I/O. BIAS2 input for LCD.
LCDBIAS3	78	I	Analog	BIAS3 input for LCD.
RE3/COM0 RE3 COM0	77	I/O O	ST Analog	Digital I/O. COM0 output for LCD.
RE4/COM1 RE4 COM1	76	I/O O	ST Analog	Digital I/O. COM1 output for LCD.
RE5/COM2 RE5 COM2	75	I/O O	ST Analog	Digital I/O. COM2 output for LCD.
RE6/COM3 RE6 COM3	74	I/O O	ST Analog	Digital I/O. COM3 output for LCD.
RE7/CCP2/SEG31 RE7 CCP2 ⁽²⁾ SEG31	73	I/O I/O O	ST ST Analog	Digital I/O. Capture 2 input/Compare 2 output/PWM2 output. SEG31 output for LCD.
I = Input P = Power $I^2C^{TM} = I^2C/SN$	tt Trigger input MBus	with C		CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD) MX Configuration bit is set

TABLE 1-4:	PIC18F8XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED))

Note 1: Default assignment for CCP2 when the CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin	Buffer	Description	
Fin Name	TQFP	Туре	Туре	Description	
				PORTF is a bidirectional I/O port.	
RF1/AN6/C2OUT/SEG19 RF1 AN6 C2OUT SEG19	23	I/O I O O	ST Analog — Analog	Digital I/O. Analog Input 6. Comparator 2 output. SEG19 output for LCD.	
RF2/AN7/C1OUT/SEG20 RF2 AN7 C1OUT SEG20	18	I/O I O O	ST Analog — Analog	Digital I/O. Analog Input 7. Comparator 1 output. SEG20 output for LCD.	
RF3/AN8/SEG21/C2INB RF3 AN8 SEG21 C2INB	17	I/O I O I	ST Analog Analog Analog	Digital I/O. Analog Input 8. SEG21 output for LCD. Comparator 2 Input B.	
RF4/AN9/SEG22/C2INA RF4 AN9 SEG22 C2INA	16	I/O I O I	ST Analog Analog Analog	Digital I/O. Analog Input 9. SEG22 output for LCD. Comparator 2 Input A.	
RF5/AN10/CVREF/ SEG23/C1INB RF5 AN10 CVREF SEG23 C1INB	15	I/O I O I	ST Analog Analog Analog Analog	Digital I/O. Analog Input 10. Comparator reference voltage output. SEG23 output for LCD. Comparator 1 Input B.	
RF6/AN11/SEG24/C1INA RF6 AN11 SEG24 C1INA	14	I/O I O I	ST Analog Analog Analog	Digital I/O. Analog Input 11. SEG24 output for LCD. Comparator 1 Input A.	
RF7/AN5/SS/SEG25 RF7 AN5 SS SEG25	13	I/O O I O	ST Analog TTL Analog	Digital I/O. Analog Input 5. SPI slave select input. SEG25 output for LCD.	
Legend:TTL= TTL compatible inputCMOS= CMOS compatible input or outputST= Schmitt Trigger input with CMOS levelsAnalog= Analog inputI= InputO= OutputP= PowerOD= Open-Drain (no P diode to VDD) $l^2C^{TM} = l^2C/SMBus$ II					

TABLE 1-4: PIC18F8XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)

Note 1: Default assignment for CCP2 when the CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin	n Buffer	Description
Pin Name	TQFP	Туре	Туре	Description
				PORTG is a bidirectional I/O port.
RG0/LCDBIAS0 RG0 LCDBIAS0	5	I/O I	ST Analog	Digital I/O. BIAS0 input for LCD.
RG1/TX2/CK2 RG1 TX2 CK2	6	I/O O I/O	ST — ST	Digital I/O. AUSART asynchronous transmit. AUSART synchronous clock (see related RX2/DT2).
RG2/RX2/DT2/VLCAP1 RG2 RX2 DT2 VLCAP1	7	I/O I I/O I	ST ST ST Analog	Digital I/O. AUSART asynchronous receive. AUSART synchronous data (see related TX2/CK2). LCD charge pump capacitor input.
RG3/VLCAP2 RG3 VLCAP2	8	I/O I	ST Analog	Digital I/O. LCD charge pump capacitor input.
RG4/SEG26/RTCC RG4 SEG26 RTCC	10	I/O O O	ST Analog —	Digital I/O. SEG26 output for LCD. RTCC output.
	/Bus			CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

Note 1: Default assignment for CCP2 when the CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin		Description		
	TQFP	Туре		Description		
				PORTH is a bidirectional I/O port.		
RH0/SEG47 RH0 SEG47	79	I/O O	ST Analog	Digital I/O. SEG47 output for LCD.		
RH1/SEG46 RH1 SEG46	80	I/O O	ST Analog	Digital I/O. SEG46 output for LCD.		
RH2/SEG45 RH2 SEG45	1	I/O O	ST Analog	Digital I/O. SEG45 output for LCD.		
RH3/SEG44 RH3 SEG44	2	I/O O	ST Analog	Digital I/O. SEG44 output for LCD.		
RH4/SEG40 RH4 SEG40	22	I/O O	ST Analog	Digital I/O. SEG40 output for LCD.		
RH5/SEG41 RH5 SEG41	21	I/O O	ST Analog	Digital I/O. SEG41 output for LCD.		
RH6/SEG42 RH6 SEG42	20	I/O O	ST Analog	Digital I/O. SEG42 output for LCD.		
RH7/SEG43 RH7 SEG43	19	I/O O	ST Analog	Digital I/O. SEG43 output for LCD.		
Legend:TTL=TTL compatible inputCMOS=CMOS compatible input or outputST=Schmitt Trigger input with CMOS levelsAnalog=Analog inputI=InputO=Output						

TABLE 1-4: PIC18F8XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED)

Р = Power

 $I^2C^{TM} = I^2C/SMBus$

OD = Open-Drain (no P diode to VDD)

Note 1: Default assignment for CCP2 when the CCP2MX Configuration bit is set.

Pin Name	Pin Number	Pin	Buffer	Description		
Fill Name	TQFP	Туре Туре		Description		
				PORTJ is a bidirectional I/O port.		
RJ0	62	I/O	ST	Digital I/O.		
RJ1/SEG33 RJ1 SEG33	61	I/O O	ST Analog	Digital I/O. g SEG33 output for LCD.		
RJ2/SEG34 RJ2 SEG34	60	I/O O	ST Analog	Digital I/O. SEG34 output for LCD.		
RJ3/SEG35 RJ3 SEG35	59	I/O O	ST Analog	Digital I/O. SEG35 output for LCD.		
RJ4/SEG39 RJ4 SEG39	39	I/O O	ST Analog	Digital I/O. SEG39 output for LCD.		
RJ5/SEG38 RJ5 SEG38	40	I/O O	ST Analog	Digital I/O SEG38 output for LCD.		
RJ6/SEG37 RJ6 SEG37	41	I/O O	ST Analog	Digital I/O. SEG37 output for LCD.		
RJ7/SEG36 RJ7 SEG36	42	I/O O	ST Analog	Digital I/O. g SEG36 output for LCD.		
Vss	11, 31, 51, 70	Р	—	Ground reference for logic and I/O pins.		
Vdd	32, 48, 71	Р		Positive supply for logic and I/O pins.		
AVss	26	Р	_	Ground reference for analog modules.		
AVDD	25	Р	—	Positive supply for analog modules.		
ENVREG	24	Ι	ST	Enable for on-chip voltage regulator.		
VDDCORE/VCAP VDDCORE	12	Ρ	_	Core logic power or external filter capacitor connection. Positive supply for microcontroller core logic (regulator disabled).		
VCAP		Р	—	External filter capacitor connection (regulator enabled).		
		with C	MOS leve	CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)		

TABLE 1-4:	PIC18F8XJ90 PINOUT I/O DESCRIPTIONS (CONTINUED))

Note 1: Default assignment for CCP2 when the CCP2MX Configuration bit is set.

PIC18F87J90 FAMILY

NOTES:

2.0 GUIDELINES FOR GETTING STARTED WITH PIC18FJ MICROCONTROLLERS

2.1 Basic Connection Requirements

Getting started with the PIC18F87J90 family family of 8-bit microcontrollers requires attention to a minimal set of device pin connections before proceeding with development.

The following pins must always be connected:

- All VDD and Vss pins (see Section 2.2 "Power Supply Pins")
- All AVDD and AVss pins, regardless of whether or not the analog device features are used (see Section 2.2 "Power Supply Pins")
- MCLR pin (see Section 2.3 "Master Clear (MCLR) Pin")
- ENVREG (if implemented) and VCAP/VDDCORE pins (see Section 2.4 "Voltage Regulator Pins (ENVREG and VCAP/VDDCORE)")

These pins must also be connected if they are being used in the end application:

- PGC/PGD pins used for In-Circuit Serial Programming[™] (ICSP[™]) and debugging purposes (see **Section 2.5 "ICSP Pins**")
- OSCI and OSCO pins when an external oscillator source is used

(see Section 2.6 "External Oscillator Pins")

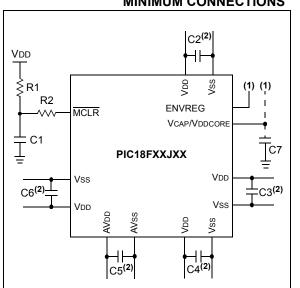
Additionally, the following pins may be required:

• VREF+/VREF- pins are used when external voltage reference for analog modules is implemented

Note: The AVDD and AVSS pins must always be connected, regardless of whether any of the analog modules are being used.

The minimum mandatory connections are shown in Figure 2-1.

FIGURE 2-1: RECOMMENDED MINIMUM CONNECTIONS



Key (all values are recommendations):

C1 through C6: 0.1 µF, 20V ceramic

C7: 10 µF, 6.3V or greater, tantalum or ceramic

R1: 10 kΩ

R2: 100Ω to 470Ω

- Note 1: See Section 2.4 "Voltage Regulator Pins (ENVREG and VCAP/VDDCORE)" for explanation of ENVREG pin connections.
 - 2: The example shown is for a PIC18F device with five VDD/Vss and AVDD/AVss pairs. Other devices may have more or less pairs; adjust the number of decoupling capacitors appropriately.

2.2 Power Supply Pins

2.2.1 DECOUPLING CAPACITORS

The use of decoupling capacitors on every pair of power supply pins, such as VDD, VSS, AVDD and AVSS, is required.

Consider the following criteria when using decoupling capacitors:

- Value and type of capacitor: A 0.1 μ F (100 nF), 10-20V capacitor is recommended. The capacitor should be a low-ESR device, with a resonance frequency in the range of 200 MHz and higher. Ceramic capacitors are recommended.
- Placement on the printed circuit board: The decoupling capacitors should be placed as close to the pins as possible. It is recommended to place the capacitors on the same side of the board as the device. If space is constricted, the capacitor can be placed on another layer on the PCB using a via; however, ensure that the trace length from the pin to the capacitor is no greater than 0.25 inch (6 mm).
- Handling high-frequency noise: If the board is experiencing high-frequency noise (upward of tens of MHz), add a second ceramic type capacitor in parallel to the above described decoupling capacitor. The value of the second capacitor can be in the range of 0.01 μ F to 0.001 μ F. Place this second capacitor next to each primary decoupling capacitor. In high-speed circuit designs, consider implementing a decade pair of capacitances as close to the power and ground pins as possible (e.g., 0.1 μ F in parallel with 0.001 μ F).
- Maximizing performance: On the board layout from the power supply circuit, run the power and return traces to the decoupling capacitors first, and then to the device pins. This ensures that the decoupling capacitors are first in the power chain. Equally important is to keep the trace length between the capacitor and the power pins to a minimum, thereby reducing PCB trace inductance.

2.2.2 TANK CAPACITORS

On boards with power traces running longer than six inches in length, it is suggested to use a tank capacitor for integrated circuits, including microcontrollers, to supply a local power source. The value of the tank capacitor should be determined based on the trace resistance that connects the power supply source to the device, and the maximum current drawn by the device in the application. In other words, select the tank capacitor so that it meets the acceptable voltage sag at the device. Typical values range from 4.7 μF to 47 μF .

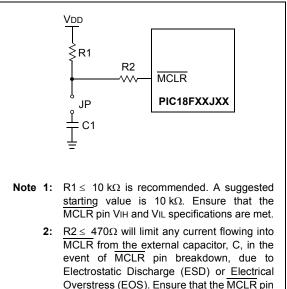
2.3 Master Clear (MCLR) Pin

The MCLR pin provides two specific device functions: Device Reset, and Device Programming and Debugging. If programming and debugging are not required in the end application, a direct connection to VDD may be all that is required. The addition of other components, to help increase the application's resistance to spurious Resets from voltage sags, may be beneficial. A typical configuration is shown in Figure 2-1. Other circuit designs may be implemented, depending on the application's requirements.

During programming and debugging, the resistance and capacitance that can be added to the pin must be considered. Device programmers and debuggers drive the $\overline{\text{MCLR}}$ pin. Consequently, specific voltage levels (VIH and VIL) and fast signal transitions must not be adversely affected. Therefore, specific values of R1 and C1 will need to be adjusted based on the application and PCB requirements. For example, it is recommended that the capacitor, C1, be isolated from the $\overline{\text{MCLR}}$ pin during programming and debugging operations by using a jumper (Figure 2-2). The jumper is replaced for normal run-time operations.

Any components associated with the $\overline{\text{MCLR}}$ pin should be placed within 0.25 inch (6 mm) of the pin.

FIGURE 2-2: EXAMPLE OF MCLR PIN CONNECTIONS



VIH and VIL specifications are met.

2.4 Voltage Regulator Pins (ENVREG and VCAP/VDDCORE)

The on-chip voltage regulator enable pin, ENVREG, must always be connected directly to either a supply voltage or to ground. Tying ENVREG to VDD enables the regulator, while tying it to ground disables the regulator. Refer to **Section 25.3 "On-Chip Voltage Regulator"** for details on connecting and using the on-chip regulator.

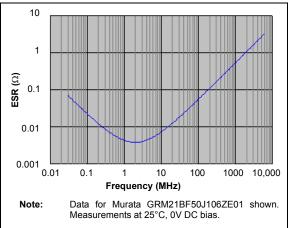
When the regulator is enabled, a low-ESR (< 5 Ω) capacitor is required on the VCAP/VDDCORE pin to stabilize the voltage regulator output voltage. The VCAP/VDDCORE pin must not be connected to VDD and must use a capacitor of 10 μ F connected to ground. The type can be ceramic or tantalum. A suitable example is the Murata GRM21BF50J106ZE01 (10 μ F, 6.3V) or equivalent. Designers may use Figure 2-3 to evaluate ESR equivalence of candidate devices.

It is recommended that the trace length not exceed 0.25 inch (6 mm). Refer to **Section 28.0 "Electrical Characteristics"** for additional information.

When the regulator is disabled, the VCAP/VDDCORE pin must be tied to a voltage supply at the VDDCORE level. Refer to **Section 28.0 "Electrical Characteristics"** for information on VDD and VDDCORE.

Note that the "LF" versions of some low pin count PIC18FJ parts (e.g., the PIC18LF45J10) do not have the ENVREG pin. These devices are provided with the voltage regulator permanently disabled; they must always be provided with a supply voltage on the VDDCORE pin.





2.5 ICSP Pins

The PGC and PGD pins are used for In-Circuit Serial ProgrammingTM (ICSPTM) and debugging purposes. It is recommended to keep the trace length between the ICSP connector and the ICSP pins on the device as short as possible. If the ICSP connector is expected to experience an ESD event, a series resistor is recommended, with the value in the range of a few tens of ohms, not to exceed 100 Ω .

Pull-up resistors, series diodes, and capacitors on the PGC and PGD pins are not recommended as they will interfere with the programmer/debugger communications to the device. If such discrete components are an application requirement, they should be removed from the circuit during programming and debugging. Alternatively, refer to the AC/DC characteristics and timing requirements information in the respective device Flash programming specification for information on capacitive loading limits, and pin input voltage high (VIH) and input low (VIL) requirements.

For device emulation, ensure that the "Communication Channel Select" (i.e., PGCx/PGDx pins) programmed into the device matches the physical connections for the ICSP to the Microchip debugger/emulator tool.

For more information on available Microchip development tools connection requirements, refer to **Section 27.0 "Development Support"**.

PIC18F87J90 FAMILY

2.6 External Oscillator Pins

Many microcontrollers have options for at least two oscillators: a high-frequency primary oscillator and a low-frequency secondary oscillator (refer to **Section 3.0 "Oscillator Configurations"** for details).

The oscillator circuit should be placed on the same side of the board as the device. Place the oscillator circuit close to the respective oscillator pins with no more than 0.5 inch (12 mm) between the circuit components and the pins. The load capacitors should be placed next to the oscillator itself, on the same side of the board.

Use a grounded copper pour around the oscillator circuit to isolate it from surrounding circuits. The grounded copper pour should be routed directly to the MCU ground. Do not run any signal traces or power traces inside the ground pour. Also, if using a two-sided board, avoid any traces on the other side of the board where the crystal is placed.

Layout suggestions are shown in Figure 2-4. In-line packages may be handled with a single-sided layout that completely encompasses the oscillator pins. With fine-pitch packages, it is not always possible to completely surround the pins and components. A suitable solution is to tie the broken guard sections to a mirrored ground layer. In all cases, the guard trace(s) must be returned to ground.

In planning the application's routing and I/O assignments, ensure that adjacent port pins and other signals in close proximity to the oscillator are benign (i.e., free of high frequencies, short rise and fall times, and other similar noise).

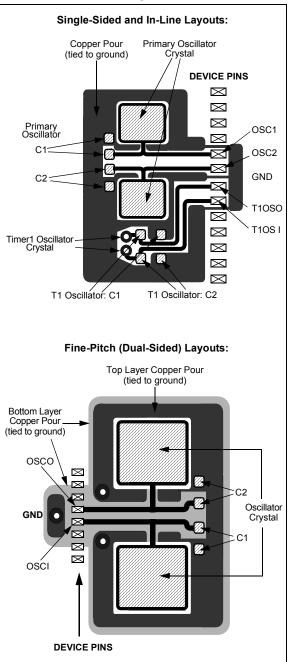
For additional information and design guidance on oscillator circuits, please refer to these Microchip Application Notes, available at the corporate web site (www.microchip.com):

- AN826, "Crystal Oscillator Basics and Crystal Selection for rfPIC™ and PICmicro[®] Devices"
- AN849, "Basic PICmicro[®] Oscillator Design"
- AN943, "Practical PICmicro[®] Oscillator Analysis and Design"
- AN949, "Making Your Oscillator Work"

2.7 Unused I/Os

Unused I/O pins should be configured as outputs and driven to a logic low state. Alternatively, connect a 1 k Ω to 10 k Ω resistor to Vss on unused pins and drive the output to logic low.

FIGURE 2-4: SUGGESTED PLACEMENT OF THE OSCILLATOR CIRCUIT



3.0 OSCILLATOR CONFIGURATIONS

3.1 Oscillator Types

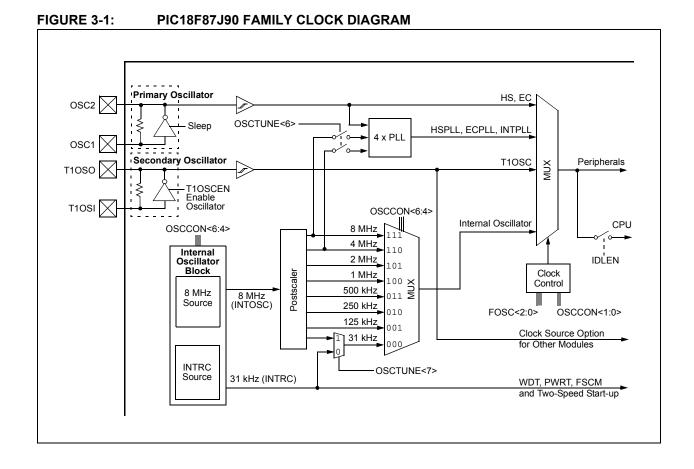
The PIC18F87J90 family of devices can be operated in eight different oscillator modes:

- 1. ECPLL OSC1/OSC2 as primary; ECPLL oscillator with PLL enabled, CLKO on RA6
- 2. EC OSC1/OSC2 as primary; external clock with Fosc/4 output
- 3. HSPLL OSC1/OSC2 as primary; high-speed crystal/resonator with software PLL control
- 4. HS OSC1/OSC2 as primary; high-speed crystal/resonator
- 5. INTPLL1 Internal oscillator block with software PLL control, Fosc/4 output on RA6 and I/O on RA7
- 6. INTIO1 Internal oscillator block with Fosc/4 output on RA6 and I/O on RA7
- 7. INTPLL2 Internal oscillator block with software PLL control and I/O on RA6 and RA7
- 8. INTIO2 Internal oscillator block with I/O on RA6 and RA7

All of these modes are selected by the user by programming the FOSC<2:0> Configuration bits.

In addition, PIC18F87J90 family devices can switch between different clock sources, either under software control or automatically under certain conditions. This allows for additional power savings by managing device clock speed in real time without resetting the application.

The clock sources for the PIC18F87J90 family of devices are shown in Figure 3-1.



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3.2 **Control Registers**

The OSCCON register (Register 3-1) controls the main aspects of the device clock's operation. It selects the oscillator type to be used, which of the power-managed modes to invoke and the output frequency of the INTOSC source. It also provides status on the oscillators. The OSCTUNE register (Register 3-2) controls the tuning and operation of the internal oscillator block. It also implements the PLLEN bits which control the operation of the Phase Locked Loop (PLL) (see Section 3.4.3 "PLL Frequency Multiplier").

OSCCON: OSCILLATOR CONTROL REGISTER REGISTER 3-1:

R/W-0	R/W-1	R/W-1	R/W-0	R ⁽¹⁾	R-0	R/W-0	R/W-0
IDLEN	IRCF2 ⁽²⁾	IRCF1 ⁽²⁾	IRCF0 ⁽²⁾	OSTS	IOFS	SCS1 ⁽⁴⁾	SCS0 ⁽⁴⁾
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **IDLEN:** Idle Enable bit 1 = Device enters an Idle mode when a SLEEP instruction is executed 0 = Device enters Sleep mode when a SLEEP instruction is executed IRCF<2:0>: INTOSC Source Frequency Select bits⁽²⁾ bit 6-4 111 = 8 MHz (INTOSC drives clock directly) 110 = 4 MHz (default) 101 = 2 MHz 100 = 1 MHz 011 = 500 kHz 010 = 250 kHz 001 = 125 kHz 000 = 31 kHz (from either INTOSC/256 or INTRC)⁽³⁾ OSTS: Oscillator Start-up Timer Time-out Status bit⁽¹⁾ bit 3 1 = Oscillator Start-up Timer (OST) time-out has expired; primary oscillator is running 0 = Oscillator Start-up Timer (OST) time-out is running; primary oscillator is not ready bit 2 **IOFS:** INTOSC Frequency Stable bit 1 = Fast RC oscillator frequency is stable 0 = Fast RC oscillator frequency is not stable SCS<1:0>: System Clock Select bits⁽⁴⁾ bit 1-0 11 = Internal oscillator block 10 = Primary oscillator 01 = Timer1 oscillator 00 = Default primary oscillator (as defined by the FOSC<2:0> Configuration bits) Note 1: Reset state depends on the state of the IESO Configuration bit. 2: Modifying these bits will cause an immediate clock frequency switch if the internal oscillator is providing

- - the device clocks. 3: Source selected by the INTSRC bit (OSCTUNE<7>), see text.
 - Modifying these bits will cause an immediate clock source switch. 4:

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
INTSRC	PLLEN	TUN5	TUN4	TUN3	TUN2	TUN1	TUN0
bit 7							bit (
Legend:							
R = Readable	e bit	W = Writable	bit	U = Unimplen	nented bit, read	d as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkn	iown
bit 7	bit 7INTSRC: Internal Oscillator Low-Frequency Source Select bit1 = 31.25 kHz device clock derived from 8 MHz INTOSC source (divide-by-256 enabled)0 = 31 kHz device clock derived from the INTRC 31 kHz oscillator						
bit 6	PLLEN: Frequence 1 = PLL enable 0 = PLL disate		r PLL Enable	bit			
bit 5-0	011111 = Ma • 0000001 000000 = Ce 111111	iximum frequer • •	ncy . Fast RC osci	Frequency Tunii		ed frequency.	

REGISTER 3-2: OSCTUNE: OSCILLATOR TUNING REGISTER

3.3 Clock Sources and Oscillator Switching

Essentially, PIC18F87J90 family devices have three independent clock sources:

- · Primary oscillators
- Secondary oscillators
- Internal oscillator

The **primary oscillators** can be thought of as the main device oscillators. These are any external oscillators connected to the OSC1 and OSC2 pins, and include the External Crystal and Resonator modes and the External Clock modes. If selected by the FOSC<2:0> Configuration bits, the internal oscillator block (either the 31 kHz INTRC or the 8 MHz INTOSC source) may be considered a primary oscillator. The particular mode is defined by the FOSC Configuration bits. The details of these modes are covered in **Section 3.4 "External Oscillator Modes"**.

The **secondary oscillators** are external clock sources that are not connected to the OSC1 or OSC2 pins. These sources may continue to operate even after the

controller is placed in a power-managed mode. PIC18F87J90 family devices offer the Timer1 oscillator as a secondary oscillator source. This oscillator, in all power-managed modes, is often the time base for functions such as a Real-Time Clock (RTC). The Timer1 oscillator is discussed in greater detail in **Section 12.0 "Timer1 Module"**.

In addition to being a primary clock source in some circumstances, the **internal oscillator** is available as a power-managed mode clock source. The INTRC source is also used as the clock source for several special features, such as the WDT and Fail-Safe Clock Monitor. The internal oscillator block is discussed in more detail in **Section 3.5** "Internal Oscillator **Block**".

The PIC18F87J90 family includes features that allow the device clock source to be switched from the main oscillator, chosen by device configuration, to one of the alternate clock sources. When an alternate clock source is enabled, various power-managed operating modes are available.

3.3.1 CLOCK SOURCE SELECTION

The System Clock Select bits, SCS<1:0> (OSCCON<1:0>), select the clock source. The available clock sources are the primary clock defined by the FOSC<2:0> Configuration bits, the secondary clock (Timer1 oscillator) and the internal oscillator. The clock source changes after one or more of the bits is written to, following a brief clock transition interval.

The OSTS (OSCCON<3>) and T1RUN (T1CON<6>) bits indicate which clock source is currently providing the device clock. The OSTS bit indicates that the Oscillator Start-up Timer (OST) has timed out and the primary clock is providing the device clock in Primary Clock modes. The T1RUN bit indicates when the Timer1 oscillator is providing the device clock in Secondary Clock modes. In power-managed modes, only one of these bits will be set at any time. If neither of these bits is set, the INTRC is providing the clock, or the internal oscillator has just started and is not yet stable.

The IDLEN bit determines if the device goes into Sleep mode or one of the Idle modes when the ${\tt SLEEP}$ instruction is executed.

The use of the flag and control bits in the OSCCON register is discussed in more detail in **Section 4.0** "Power-Managed Modes".

- Note 1: The Timer1 oscillator must be enabled to select the secondary clock source. The Timer1 oscillator is enabled by setting the T1OSCEN bit in the Timer1 Control register (T1CON<3>). If the Timer1 oscillator is not enabled, then any attempt to select a secondary clock source when executing a SLEEP instruction will be ignored.
 - 2: It is recommended that the Timer1 oscillator be operating and stable before executing the SLEEP instruction or a very long delay may occur while the Timer1 oscillator starts.

3.3.1.1 System Clock Selection and Device Resets

Since the SCS bits are cleared on all forms of Reset, this means the primary oscillator defined by the FOSC<2:0> Configuration bits is used as the primary clock source on device Resets. This could either be the internal oscillator block by itself or one of the other primary clock source (HS, EC, HSPLL, ECPLL1/2 or INTPLL1/2).

In those cases when the internal oscillator block, without PLL, is the default clock on Reset, the Fast RC oscillator (INTOSC) will be used as the device clock source. It will initially start at 1 MHz; the postscaler selection that corresponds to the Reset value of the IRCF<2:0> bits ('100').

Regardless of which primary oscillator is selected, INTRC will always be enabled on device power-up. It serves as the clock source until the device has loaded its configuration values from memory. It is at this point that the FOSC Configuration bits are read and the oscillator selection of the operational mode is made.

Note that either the primary clock source, or the internal oscillator, will have two bit setting options for the possible values of the SCS<1:0> bits at any given time.

3.3.2 OSCILLATOR TRANSITIONS

PIC18F87J90 family devices contain circuitry to prevent clock "glitches" when switching between clock sources. A short pause in the device clock occurs during the clock switch. The length of this pause is the sum of two cycles of the old clock source and three to four cycles of the new clock source. This formula assumes that the new clock source is stable.

Clock transitions are discussed in greater detail in **Section 4.1.2 "Entering Power-Managed Modes"**.

3.4 External Oscillator Modes

3.4.1 CRYSTAL OSCILLATOR/CERAMIC RESONATORS (HS MODES)

In HS or HSPLL Oscillator modes, a crystal or ceramic resonator is connected to the OSC1 and OSC2 pins to establish oscillation. Figure 3-2 shows the pin connections.

The oscillator design requires the use of a crystal rated for parallel resonant operation.

Note: Use of a crystal rated for series resonant operation may give a frequency out of the crystal manufacturer's specifications.

TABLE 3-1:CAPACITOR SELECTION FOR
CERAMIC RESONATORS

-	Typical Capac	itor Values Use	ed:
Mode	Freq.	OSC1	OSC2
HS	8.0 MHz 16.0 MHz	27 pF 22 pF	27 pF 22 pF

Capacitor values are for design guidance only.

Different capacitor values may be required to produce acceptable oscillator operation. The user should test the performance of the oscillator over the expected VDD and temperature range for the application. Refer to the following application notes for oscillator specific information:

- AN588, "PIC[®] Microcontroller Oscillator Design Guide"
- AN826, "Crystal Oscillator Basics and Crystal Selection for rfPIC[®] and PIC[®] Devices"
- AN849, "Basic PIC[®] Oscillator Design"
- AN943, "Practical PIC[®] Oscillator Analysis and Design"
- AN949, "Making Your Oscillator Work"

See the notes following Table 3-2 for additional information.

TABLE 3-2:CAPACITOR SELECTION FOR
CRYSTAL OSCILLATOR

Osc Type	Crystal Freq.	Typical Capa Tes	
	Fled.	C1	C2
HS	4 MHz	27 pF	27 pF
	8 MHz	22 pF	22 pF
	20 MHz	15 pF	15 pF

Capacitor values are for design guidance only.

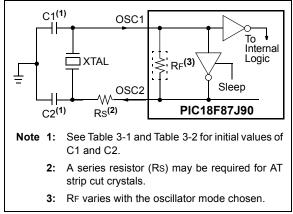
Different capacitor values may be required to produce acceptable oscillator operation. The user should test the performance of the oscillator over the expected VDD and temperature range for the application.

Refer to the Microchip application notes cited in Table 3-1 for oscillator specific information. Also see the notes following this table for additional information.

- Note 1: Higher capacitance increases the stability of oscillator but also increases the start-up time.
 - 2: Since each resonator/crystal has its own characteristics, the user should consult the resonator/crystal manufacturer for appropriate values of external components.
 - **3:** Rs may be required to avoid overdriving crystals with low drive level specification.
 - **4:** Always verify oscillator performance over the VDD and temperature range that is expected for the application.

FIGURE 3-2:

CRYSTAL/CERAMIC RESONATOR OPERATION (HS OR HSPLL CONFIGURATION)

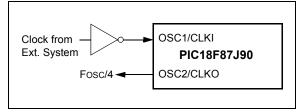


3.4.2 EXTERNAL CLOCK INPUT (EC MODES)

The EC and ECPLL Oscillator modes require an external clock source to be connected to the OSC1 pin. There is no oscillator start-up time required after a Power-on Reset or after an exit from Sleep mode.

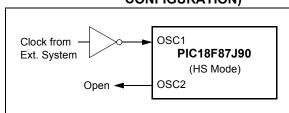
In the EC Oscillator mode, the oscillator frequency divided by 4 is available on the OSC2 pin. This signal may be used for test purposes or to synchronize other logic. Figure 3-3 shows the pin connections for the EC Oscillator mode.

FIGURE 3-3: EXTERNAL CLOCK INPUT OPERATION (EC CONFIGURATION)



An external clock source may also be connected to the OSC1 pin in the HS mode, as shown in Figure 3-4. In this configuration, the divide-by-4 output on OSC2 is not available. Current consumption in this configuration will be somewhat higher than EC mode, as the internal oscillator's feedback circuitry will be enabled (in EC mode, the feedback circuit is disabled).

FIGURE 3-4: EXTERNAL CLOCK INPUT OPERATION (HS OSC CONFIGURATION)



3.4.3 PLL FREQUENCY MULTIPLIER

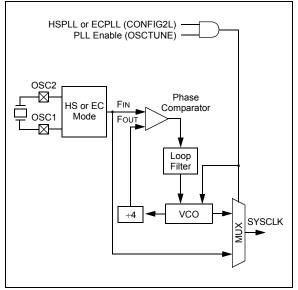
A Phase Locked Loop (PLL) circuit is provided as an option for users who want to use a lower frequency oscillator circuit, or to clock the device up to its highest rated frequency from a crystal oscillator. This may be useful for customers who are concerned with EMI due to high-frequency crystals, or users who require higher clock speeds from an internal oscillator.

3.4.3.1 HSPLL and ECPLL Modes

The HSPLL and ECPLL modes provide the ability to selectively run the device at 4 times the external oscillating source to produce frequencies of up to 40 MHz.

The PLL is enabled by programming the FOSC<2:0> Configuration bits to either '111' (for ECPLL) or '101' (for HSPLL). In addition, the PLLEN bit (OSCTUNE<6>) must also be set. Clearing PLLEN disables the PLL, regardless of the chosen oscillator configuration. It also allows additional flexibility for controlling the application's clock speed in software.

FIGURE 3-5: PLL BLOCK DIAGRAM



3.4.3.2 PLL and INTOSC

The PLL is also available to the internal oscillator block when the internal oscillator block is configured as the primary clock source. In this configuration, the PLL is enabled in software and generates a clock output of up to 32 MHz. The operation of INTOSC with the PLL is described in **Section 3.5.2 "INTPLL Modes"**.

3.5 Internal Oscillator Block

The PIC18F87J90 family of devices includes an internal oscillator block which generates two different clock signals; either can be used as the micro-controller's clock source. This may eliminate the need for an external oscillator circuit on the OSC1 and/or OSC2 pins.

The main output is the Fast RC oscillator or INTOSC, an 8 MHz clock source which can be used to directly drive the device clock. It also drives a postscaler, which can provide a range of clock frequencies from 31 kHz to 4 MHz. INTOSC is enabled when a clock frequency from 125 kHz to 8 MHz is selected. The INTOSC output can also be enabled when 31 kHz is selected, depending on the INTSRC bit (OSCTUNE<7>).

The other clock source is the Internal RC oscillator (INTRC), which provides a nominal 31 kHz output. INTRC is enabled if it is selected as the device clock source; it is also enabled automatically when any of the following are enabled:

- Power-up Timer
- Fail-Safe Clock Monitor
- Watchdog Timer
- Two-Speed Start-up

These features are discussed in greater detail in **Section 25.0 "Special Features of the CPU"**.

The clock source frequency (INTOSC direct, INTOSC with postscaler or INTRC direct) is selected by configuring the IRCF bits of the OSCCON register. The default frequency on device Resets is 4 MHz.

3.5.1 INTIO MODES

Using the internal oscillator as the clock source eliminates the need for up to two external oscillator pins, which can then be used for digital I/O. Two distinct oscillator configurations, which are determined by the FOSC Configuration bits, are available:

- In INTIO1 mode, the OSC2 pin outputs Fosc/4, while OSC1 functions as RA7 (see Figure 3-6) for digital input and output.
- In INTIO2 mode, OSC1 functions as RA7 and OSC2 functions as RA6 (see Figure 3-7), both for digital input and output.

FIGURE 3-6: INTIO1 OSCILLATOR MODE

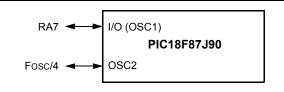


FIGURE 3-7: INTIO2 OSCILLATOR MODE



3.5.2 INTPLL MODES

The 4x Phase Locked Loop (PLL) can be used with the internal oscillator block to produce faster device clock speeds than are normally possible with the internal oscillator sources. When enabled, the PLL produces a clock speed of 16 MHz or 32 MHz.

PLL operation is controlled through software. The control bit, PLLEN (OSCTUNE<6>), is used to enable or disable its operation. The PLL is available only to INTOSC when the device is configured to use one of the INTPLL modes as the primary clock source (FOSC<2:0> = 011 or 001). Additionally, the PLL will only function when the selected output frequency is either 4 MHz or 8 MHz (OSCCON<6:4> = 111 or 110).

Like the INTIO modes, there are two distinct INTPLL modes available:

- In INTPLL1 mode, the OSC2 pin outputs Fosc/4, while OSC1 functions as RA7 for digital input and output. Externally, this is identical in appearance to INTIO1 (Figure 3-6).
- In INTPLL2 mode, OSC1 functions as RA7 and OSC2 functions as RA6, both for digital input and output. Externally, this is identical to INTIO2 (Figure 3-7).

3.5.3 INTERNAL OSCILLATOR OUTPUT FREQUENCY AND TUNING

The internal oscillator block is calibrated at the factory to produce an INTOSC output frequency of 8 MHz. It can be adjusted in the user's application by writing to TUN<5:0> (OSCTUNE<5:0>) in the OSCTUNE register (Register 3-2).

When the OSCTUNE register is modified, the INTOSC frequency will begin shifting to the new frequency. The oscillator will stabilize within 1 ms. Code execution continues during this shift and there is no indication that the shift has occurred.

The INTRC oscillator operates independently of the INTOSC source. Any changes in INTOSC across voltage and temperature are not necessarily reflected by changes in INTRC or vice versa. The frequency of INTRC is not affected by OSCTUNE.

3.5.4 INTOSC FREQUENCY DRIFT

The INTOSC frequency may drift as VDD or temperature changes, and can affect the controller operation in a variety of ways. It is possible to adjust the INTOSC frequency by modifying the value in the OSCTUNE register. Depending on the device, this may have no effect on the INTRC clock source frequency.

Tuning INTOSC requires knowing when to make the adjustment, in which direction it should be made, and in some cases, how large a change is needed. Three compensation techniques are shown here.

3.5.4.1 Compensating with the EUSART

An adjustment may be required when the EUSART begins to generate framing errors or receives data with errors while in Asynchronous mode. Framing errors indicate that the device clock frequency is too high. To adjust for this, decrement the value in OSCTUNE to reduce the clock frequency. On the other hand, errors in data may suggest that the clock speed is too low. To compensate, increment OSCTUNE to increase the clock frequency.

3.5.4.2 Compensating with the Timers

This technique compares device clock speed to some reference clock. Two timers may be used; one timer is clocked by the peripheral clock, while the other is clocked by a fixed reference source, such as the Timer1 oscillator.

Both timers are cleared, but the timer clocked by the reference generates interrupts. When an interrupt occurs, the internally clocked timer is read and both timers are cleared. If the internally clocked timer value is much greater than expected, then the internal oscillator block is running too fast. To adjust for this, decrement the OSCTUNE register.

3.5.4.3 Compensating with the CCP Module in Capture Mode

A CCP module can use free-running Timer1 (or Timer3), clocked by the internal oscillator block and an external event with a known period (i.e., AC power frequency). The time of the first event is captured in the CCPRxH:CCPRxL registers and is recorded for use later. When the second event causes a capture, the time of the first event is subtracted from the time of the second event. Since the period of the external event is known, the time difference between events can be calculated.

If the measured time is much greater than the calculated time, the internal oscillator block is running too fast. To compensate, decrement the OSCTUNE register. If the measured time is much less than the calculated time, the internal oscillator block is running too slow. To compensate, increment the OSCTUNE register.

3.6 Effects of Power-Managed Modes on the Various Clock Sources

When PRI_IDLE mode is selected, the designated primary oscillator continues to run without interruption. For all other power-managed modes, the oscillator using the OSC1 pin is disabled. The OSC1 pin (and OSC2 pin if used by the oscillator) will stop oscillating.

In Secondary Clock modes (SEC_RUN and SEC_IDLE), the Timer1 oscillator is operating and providing the device clock. The Timer1 oscillator may also run in all power-managed modes if required to clock Timer1 or Timer3.

In RC_RUN and RC_IDLE modes, the internal oscillator provides the device clock source. The 31 kHz INTRC output can be used directly to provide the clock and may be enabled to support various special features, regardless of the power-managed mode (see Section 25.2 "Watchdog Timer (WDT)" through Section 25.5 "Fail-Safe Clock Monitor" for more information on WDT, Fail-Safe Clock Monitor and Two-Speed Start-up).

If Sleep mode is selected, all clock sources are stopped. Since all the transistor switching currents have been stopped, Sleep mode achieves the lowest current consumption of the device (only leakage currents).

Enabling any on-chip feature that will operate during Sleep will increase the current consumed during Sleep. The INTRC is required to support WDT operation. The Timer1 oscillator may be operating to support a Real-Time Clock (RTC). Other features may be operating that do not require a device clock source (i.e., MSSP slave, INTx pins and others). Peripherals that may add significant current consumption are listed in Section 28.2 "DC Characteristics: Power-Down and Supply Current PIC18F87J90 Family (Industrial)".

3.7 Power-up Delays

Power-up delays are controlled by two timers, so that no external Reset circuitry is required for most applications. The delays ensure that the device is kept in Reset until the device power supply is stable under normal circumstances and the primary clock is operating and stable. For additional information on power-up delays, see **Section 5.6 "Power-up Timer (PWRT)**".

The first timer is the Power-up Timer (PWRT), which provides a fixed delay on power-up (parameter 33, Table 28-11); it is always enabled.

The second timer is the Oscillator Start-up Timer (OST), intended to keep the chip in Reset until the crystal oscillator is stable (HS modes). The OST does this by counting 1024 oscillator cycles before allowing the oscillator to clock the device.

There is a delay of interval, TCSD (parameter 38, Table 28-11), following POR, while the controller becomes ready to execute instructions.

Oscillator Mode	OSC1 Pin	OSC2 Pin
EC, ECPLL	Floating, pulled by external clock	At logic low (clock/4 output)
HS, HSPLL	Feedback inverter disabled at quiescent voltage level	Feedback inverter disabled at quiescent voltage level
INTOSC, INTPLL1/2	I/O pin RA6, direction controlled by TRISA<6>	I/O pin RA6, direction controlled by TRISA<7>

 TABLE 3-3:
 OSC1 AND OSC2 PIN STATES IN SLEEP MODE

Note: See Section 5.0 "Reset" for time-outs due to Sleep and MCLR Reset.

PIC18F87J90 FAMILY

NOTES:

4.0 POWER-MANAGED MODES

The PIC18F87J90 family devices provide the ability to manage power consumption by simply managing clocking to the CPU and the peripherals. In general, a lower clock frequency and a reduction in the number of circuits being clocked constitutes lower consumed power. For the sake of managing power in an application, there are three primary modes of operation:

- Run mode
- Idle mode
- · Sleep mode

These modes define which portions of the device are clocked and at what speed. The Run and Idle modes may use any of the three available clock sources (primary, secondary or internal oscillator block); the Sleep mode does not use a clock source.

The power-managed modes include several power-saving features offered on previous PIC[®] devices. One is the clock switching feature, offered in other PIC18 devices, allowing the controller to use the Timer1 oscillator in place of the primary oscillator. Also included is the Sleep mode, offered by all PIC devices, where all device clocks are stopped.

4.1 Selecting Power-Managed Modes

Selecting a power-managed mode requires two decisions: if the CPU is to be clocked or not and which clock source is to be used. The IDLEN bit (OSCCON<7>) controls CPU clocking, while the SCS<1:0> bits (OSCCON<1:0>) select the clock source. The individual modes, bit settings, clock sources and affected modules are summarized in Table 4-1.

4.1.1 CLOCK SOURCES

The SCS<1:0> bits allow the selection of one of three clock sources for power-managed modes. They are:

- the primary clock, as defined by the FOSC<2:0> Configuration bits
- the secondary clock (Timer1 oscillator)
- the internal oscillator

4.1.2 ENTERING POWER-MANAGED MODES

Switching from one power-managed mode to another begins by loading the OSCCON register. The SCS<1:0> bits select the clock source and determine which Run or Idle mode is to be used. Changing these bits causes an immediate switch to the new clock source, assuming that it is running. The switch may also be subject to clock transition delays. These are discussed in **Section 4.1.3 "Clock Transitions and Status Indicators"** and subsequent sections.

Entry to the power-managed Idle or Sleep modes is triggered by the execution of a SLEEP instruction. The actual mode that results depends on the status of the IDLEN bit.

Depending on the current mode and the mode being switched to, a change to a power-managed mode does not always require setting all of these bits. Many transitions may be done by changing the oscillator select bits, or changing the IDLEN bit, prior to issuing a SLEEP instruction. If the IDLEN bit is already configured correctly, it may only be necessary to perform a SLEEP instruction to switch to the desired mode.

., .= ==					
Mada	OSC	OSCCON Bits		e Clocking	Augilable Cleak and Occillator Source
Mode	IDLEN<7>(1)	SCS<1:0>	CPU Peripherals		Available Clock and Oscillator Source
Sleep	0	N/A	Off	Off	None – All clocks are disabled
PRI_RUN	N/A	10	Clocked	Clocked	Primary – HS, EC, HSPLL, ECPLL; this is the normal full-power execution mode
SEC_RUN	N/A	01	Clocked	Clocked	Secondary – Timer1 Oscillator
RC_RUN	N/A	11	Clocked	Clocked	Internal Oscillator
PRI_IDLE	1	10	Off	Clocked	Primary – HS, EC, HSPLL, ECPLL
SEC_IDLE	1	01	Off	Clocked	Secondary – Timer1 Oscillator
RC_IDLE	1	11	Off	Clocked	Internal Oscillator

TABLE 4-1: POWER-MANAGED MODES

Note 1: IDLEN reflects its value when the **SLEEP** instruction is executed.

4.1.3 CLOCK TRANSITIONS AND STATUS INDICATORS

The length of the transition between clock sources is the sum of two cycles of the old clock source and three to four cycles of the new clock source. This formula assumes that the new clock source is stable.

Two bits indicate the current clock source and its status: OSTS (OSCCON<3>) and T1RUN (T1CON<6>). In general, only one of these bits will be set while in a given power-managed mode. When the OSTS bit is set, the primary clock is providing the device clock. When the T1RUN bit is set, the Timer1 oscillator is providing the clock. If neither of these bits is set, INTRC is clocking the device.

Note: Executing a SLEEP instruction does not necessarily place the device into Sleep mode. It acts as the trigger to place the controller into either the Sleep mode, or one of the Idle modes, depending on the setting of the IDLEN bit.

4.1.4 MULTIPLE SLEEP COMMANDS

The power-managed mode that is invoked with the SLEEP instruction is determined by the setting of the IDLEN bit at the time the instruction is executed. If another SLEEP instruction is executed, the device will enter the power-managed mode specified by IDLEN at that time. If IDLEN has changed, the device will enter the new power-managed mode specified by the new setting.

4.2 Run Modes

In the Run modes, clocks to both the core and peripherals are active. The difference between these modes is the clock source.

4.2.1 PRI_RUN MODE

The PRI_RUN mode is the normal, full-power execution mode of the microcontroller. This is also the default mode upon a device Reset unless Two-Speed Start-up is enabled (see **Section 25.4 "Two-Speed Start-up"** for details). In this mode, the OSTS bit is set (see **Section 3.2 "Control Registers"**).

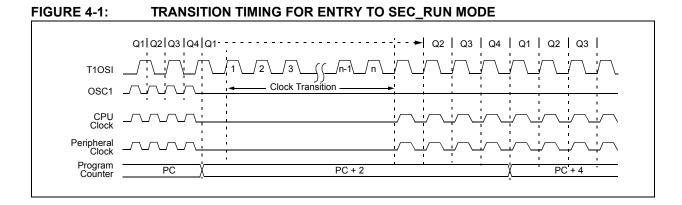
4.2.2 SEC_RUN MODE

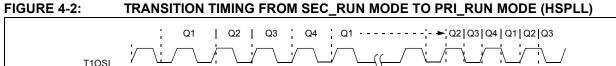
The SEC_RUN mode is the compatible mode to the "clock switching" feature offered in other PIC18 devices. In this mode, the CPU and peripherals are clocked from the Timer1 oscillator. This gives users the option of lower power consumption while still using a high-accuracy clock source.

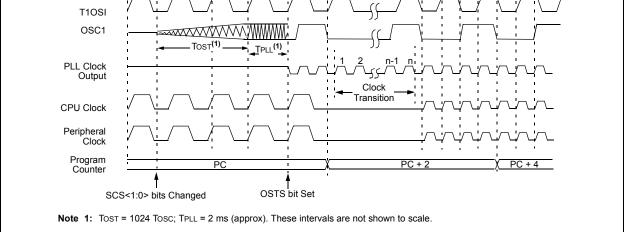
SEC_RUN mode is entered by setting the SCS<1:0> bits to '01'. The device clock source is switched to the Timer1 oscillator (see Figure 4-1), the primary oscillator is shut down, the T1RUN bit (T1CON<6>) is set and the OSTS bit is cleared.

Note: The Timer1 oscillator should already be running prior to entering SEC_RUN mode. If the T1OSCEN bit is not set when the SCS<1:0> bits are set to '01', entry to SEC_RUN mode will not occur. If the Timer1 oscillator is enabled, but not yet running, device clocks will be delayed until the oscillator has started. In such situations, initial oscillator operation is far from stable and unpredictable operation may result. On transitions from SEC_RUN mode to PRI_RUN mode, the peripherals and CPU continue to be clocked from the Timer1 oscillator while the primary clock is started. When the primary clock becomes ready, a clock switch back to the primary clock occurs (see

Figure 4-2). When the clock switch is complete, the T1RUN bit is cleared, the OSTS bit is set and the primary clock is providing the clock. The IDLEN and SCS bits are not affected by the wake-up; the Timer1 oscillator continues to run.





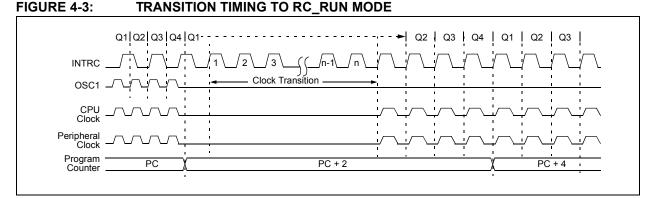


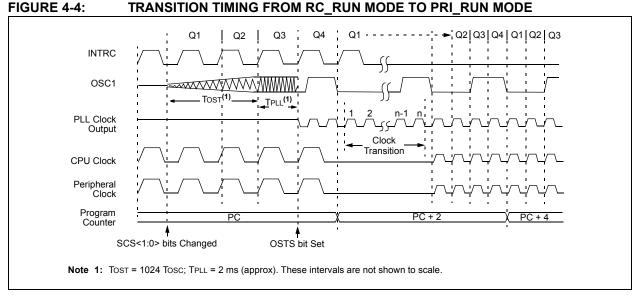
4.2.3 RC_RUN MODE

In RC_RUN mode, the CPU and peripherals are clocked from the internal oscillator; the primary clock is shut down. This mode provides the best power conservation of all the Run modes while still executing code. It works well for user applications which are not highly timing-sensitive or do not require high-speed clocks at all times.

This mode is entered by setting the SCS bits to '11'. When the clock source is switched to the INTRC (see Figure 4-3), the primary oscillator is shut down and the OSTS bit is cleared.

On transitions from RC_RUN mode to PRI_RUN mode, the device continues to be clocked from the INTRC while the primary clock is started. When the primary clock becomes ready, a clock switch to the primary clock occurs (see Figure 4-4). When the clock switch is complete, the OSTS bit is set and the primary clock is providing the device clock. The IDLEN and SCS bits are not affected by the switch. The INTRC source will continue to run if either the WDT or the Fail-Safe Clock Monitor is enabled.





4.3 Sleep Mode

The power-managed Sleep mode is identical to the legacy Sleep mode offered in all other PIC devices. It is entered by clearing the IDLEN bit (the default state on device Reset) and executing the SLEEP instruction. This shuts down the selected oscillator (Figure 4-5). All clock source status bits are cleared.

Entering the Sleep mode from any other mode does not require a clock switch. This is because no clocks are needed once the controller has entered Sleep. If the WDT is selected, the INTRC source will continue to operate. If the Timer1 oscillator is enabled, it will also continue to run.

When a wake event occurs in Sleep mode (by interrupt, Reset or WDT time-out), the device will not be clocked until the clock source, selected by the SCS<1:0> bits, becomes ready (see Figure 4-6), or it will be clocked from the internal oscillator if either the Two-Speed Start-up or the Fail-Safe Clock Monitor is enabled (see **Section 25.0 "Special Features of the CPU"**). In either case, the OSTS bit is set when the primary clock is providing the device clocks. The IDLEN and SCS bits are not affected by the wake-up.

4.4 Idle Modes

The Idle modes allow the controller's CPU to be selectively shut down while the peripherals continue to operate. Selecting a particular Idle mode allows users to further manage power consumption.

If the IDLEN bit is set to a '1' when a SLEEP instruction is executed, the peripherals will be clocked from the clock source selected using the SCS<1:0> bits; however, the CPU will not be clocked. The clock source status bits are not affected. Setting IDLEN and executing a SLEEP instruction provides a quick method of switching from a given Run mode to its corresponding Idle mode.

If the WDT is selected, the INTRC source will continue to operate. If the Timer1 oscillator is enabled, it will also continue to run.

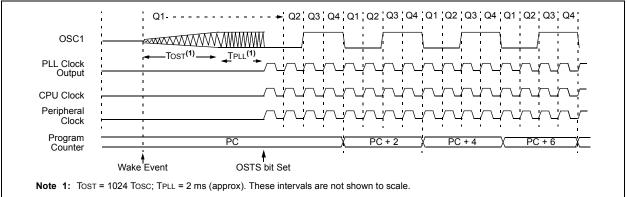
Since the CPU is not executing instructions, the only exits from any of the Idle modes are by interrupt, WDT time-out or a Reset. When a wake event occurs, CPU execution is delayed by an interval of TCSD (parameter 38, Table 28-11) while it becomes ready to execute code. When the CPU begins executing code, it resumes with the same clock source for the current Idle mode. For example, when waking from RC_IDLE mode, the internal oscillator block will clock the CPU and peripherals (in other words, RC_RUN mode). The IDLEN and SCS bits are not affected by the wake-up.

While in any Idle mode or Sleep mode, a WDT time-out will result in a WDT wake-up to the Run mode currently specified by the SCS<1:0> bits.

FIGURE 4-5: TRANSITION TIMING FOR ENTRY TO SLEEP MODE

Q1 Q2 Q3 Q4	lq1			-;	;		.;	;	
		1	1	1	1	1	1	1	1
		1	1	1	1	1		1	1
			1	1 1 1	1	- - -	;	;	
PeripheralClock		1	i 1	1	i T	1 T	1 T	1 	
Sleep	1 1 1 1 1		1	1	1	1 1	1 1	<u> </u>	ı
		1	1	1	1 1	1	1	1	1 1
Program PC	PC + 2				1	•	•	<u> </u>	





4.4.1 PRI_IDLE MODE

This mode is unique among the three low-power Idle modes, in that it does not disable the primary device clock. For timing-sensitive applications, this allows for the fastest resumption of device operation with its more accurate primary clock source, since the clock source does not have to "warm up" or transition from another oscillator.

PRI_IDLE mode is entered from PRI_RUN mode by setting the IDLEN bit and executing a SLEEP instruction. If the device is in another Run mode, set IDLEN first, then set the SCS bits to '10' and execute SLEEP. Although the CPU is disabled, the peripherals continue to be clocked from the primary clock source specified by the FOSC<1:0> Configuration bits. The OSTS bit remains set (see Figure 4-7).

When a wake event occurs, the CPU is clocked from the primary clock source. A delay of interval, TCSD, is required between the wake event and when code execution starts. This is required to allow the CPU to become ready to execute instructions. After the wake-up, the OSTS bit remains set. The IDLEN and SCS bits are not affected by the wake-up (see Figure 4-8).

4.4.2 SEC_IDLE MODE

In SEC_IDLE mode, the CPU is disabled but the peripherals continue to be clocked from the Timer1 oscillator. This mode is entered from SEC_RUN by setting the IDLEN bit and executing a SLEEP instruction. If the device is in another Run mode, set IDLEN first, then set SCS<1:0> to '01' and execute SLEEP. When the clock source is switched to the Timer1 oscillator, the primary oscillator is shut down, the OSTS bit is cleared and the T1RUN bit is set.

When a wake event occurs, the peripherals continue to be clocked from the Timer1 oscillator. After an interval of TCSD, following the wake event, the CPU begins executing code being clocked by the Timer1 oscillator. The IDLEN and SCS bits are not affected by the wake-up; the Timer1 oscillator continues to run (see Figure 4-8).

Note: The Timer1 oscillator should already be running prior to entering SEC_IDLE mode. If the T1OSCEN bit is not set when the SLEEP instruction is executed, the SLEEP instruction will be ignored and entry to SEC_IDLE mode will not occur. If the Timer1 oscillator is enabled, but not yet running, peripheral clocks will be delayed until the oscillator has started. In such situations, initial oscillator operation is far from stable and unpredictable operation may result.

FIGURE 4-7: TRANSITION TIMING FOR ENTRY TO IDLE MODE Q1 Q2 Q3 Q4 Q1 OSC1 Q1 Q2 Q3 Q4 Q1

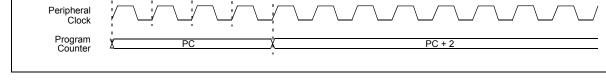
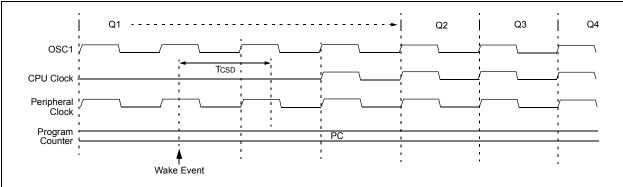


FIGURE 4-8: TRANSITION TIMING FOR WAKE FROM IDLE TO RUN MODE



CPU Clock

4.4.3 RC_IDLE MODE

In RC_IDLE mode, the CPU is disabled but the peripherals continue to be clocked from the internal oscillator. This mode allows for controllable power conservation during Idle periods.

From RC_RUN, this mode is entered by setting the IDLEN bit and executing a SLEEP instruction. If the device is in another Run mode, first set IDLEN, then clear the SCS bits and execute SLEEP. When the clock source is switched to the INTRC, the primary oscillator is shut down and the OSTS bit is cleared.

When a wake event occurs, the peripherals continue to be clocked from the INTOSC. After a delay of TCSD, following the wake event, the CPU begins executing code being clocked by the INTOSC. The IDLEN and SCS bits are not affected by the wake-up. The INTOSC source will continue to run if either the WDT or the Fail-Safe Clock Monitor is enabled.

4.5 Exiting Idle and Sleep Modes

An exit from Sleep mode, or any of the Idle modes, is triggered by an interrupt, a Reset or a WDT time-out. This section discusses the triggers that cause exits from power-managed modes. The clocking subsystem actions are discussed in each of the power-managed mode sections (see Section 4.2 "Run Modes", Section 4.3 "Sleep Mode" and Section 4.4 "Idle Modes").

4.5.1 EXIT BY INTERRUPT

Any of the available interrupt sources can cause the device to exit from an Idle mode, or the Sleep mode, to a Run mode. To enable this functionality, an interrupt source must be enabled by setting its enable bit in one of the INTCON or PIE registers. The exit sequence is initiated when the corresponding interrupt flag bit is set.

On all exits from Idle or Sleep modes by interrupt, code execution branches to the interrupt vector if the GIE/GIEH bit (INTCON<7>) is set. Otherwise, code execution continues or resumes without branching (see Section 9.0 "Interrupts").

A fixed delay of interval, TCSD, following the wake event, is required when leaving Sleep and Idle modes. This delay is required for the CPU to prepare for execution. Instruction execution resumes on the first clock cycle following this delay.

4.5.2 EXIT BY WDT TIME-OUT

A WDT time-out will cause different actions depending on which power-managed mode the device is in when the time-out occurs.

If the device is not executing code (all Idle modes and Sleep mode), the time-out will result in an exit from the power-managed mode (see Section 4.2 "Run Modes" and Section 4.3 "Sleep Mode"). If the device is executing code (all Run modes), the time-out will result in a WDT Reset (see Section 25.2 "Watchdog Timer (WDT)").

The Watchdog Timer and postscaler are cleared by one of the following events:

- executing a SLEEP or CLRWDT instruction
- the loss of a currently selected clock source (if the Fail-Safe Clock Monitor is enabled)
- 4.5.3 EXIT BY RESET

Exiting an Idle or Sleep mode by Reset automatically forces the device to run from the INTRC.

4.5.4 EXIT WITHOUT AN OSCILLATOR START-UP DELAY

Certain exits from power-managed modes do not invoke the OST at all. There are two cases:

- PRI_IDLE mode, where the primary clock source is not stopped; and
- the primary clock source is either the EC or ECPLL mode.

In these instances, the primary clock source either does not require an oscillator start-up delay, since it is already running (PRI_IDLE), or normally does not require an oscillator start-up delay (EC). However, a fixed delay of interval, TCSD, following the wake event is still required when leaving Sleep and Idle modes to allow the CPU to prepare for execution. Instruction execution resumes on the first clock cycle following this delay.

PIC18F87J90 FAMILY

NOTES:

5.0 RESET

The PIC18F87J90 family of devices differentiates between various kinds of Reset:

- Power-on Reset (POR)
- MCLR Reset during normal operation
- MCLR Reset during power-managed modes
 Watchdog Timer (WDT) Reset (during
- execution)
- Brown-out Reset (BOR)
- Configuration Mismatch (CM) Reset
- RESET Instruction
- Stack Full Reset
- Stack Underflow Reset

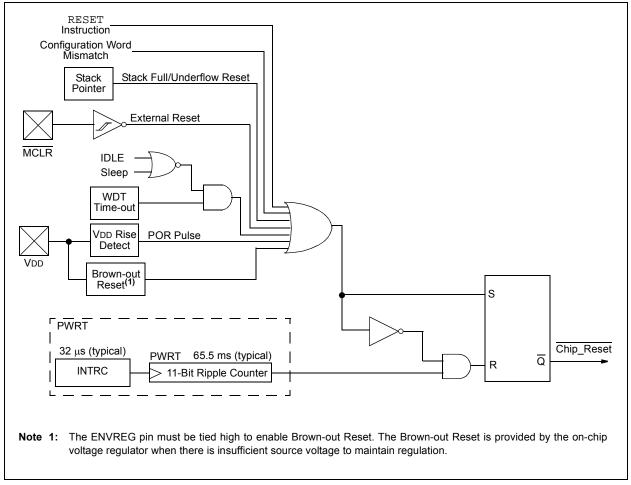
This section discusses Resets generated by MCLR, POR and BOR, and covers the operation of the various start-up timers. Stack Reset events are covered in Section 6.1.4.4 "Stack Full and Underflow Resets". WDT Resets are covered in Section 25.2 "Watchdog Timer (WDT)". A simplified block diagram of the on-chip Reset circuit is shown in Figure 5-1.

5.1 RCON Register

Device Reset events are tracked through the RCON register (Register 5-1). The lower five bits of the register indicate that a specific Reset event has occurred. In most cases, these bits can only be set by the event and must be cleared by the application after the event. The state of these flag bits, taken together, can be read to indicate the type of Reset that just occurred. This is described in more detail in **Section 5.7 "Reset State of Registers"**.

The RCON register also has a control bit for setting interrupt priority (IPEN). Interrupt priority is discussed in **Section 9.0 "Interrupts"**.

FIGURE 5-1: SIMPLIFIED BLOCK DIAGRAM OF ON-CHIP RESET CIRCUIT



PIC18F87J90 FAMILY

R/W-0	U-0	R/W-1	R/W-1	R-1	R-1	R/W-0	R/W-0
IPEN		CM	RI	TO	PD	POR	BOR
bit 7							bit
Legend:							
R = Readab	le bit	W = Writable	bit	U = Unimpler	nented bit, rea	id as '0'	
-n = Value a		'1' = Bit is set		'0' = Bit is cle		x = Bit is unkr	nown
bit 7	IPEN: Intern	upt Priority Enat	ole bit				
		priority levels or					
		priority levels of	• •	IC16XXXX Con	npatibility mod	e)	
bit 6		ented: Read as '					
bit 5	•	uration Mismatch	•				
		guration mismate				<i></i>	
	0 = A confi Reset o	guration mismat	ch has occurre	ed (must be se	t in software a	itter a Configura	ition Mismato
bit 4		nstruction Flag b	it				
		SET instruction v		ted (set by firm)	ware only)		
	0 = The RE	SET instruction	was executed			ust be set in so	oftware after
		out Reset occurs	/				
bit 3		log Time-out Fla	-				
		power-up, CLRW		or SLEEP instru	iction		
bit 2		Down Detection					
		power-up or by t	•	struction			
		execution of the					
bit 1	POR: Powe	r-on Reset Statu	s bit				
		er-on Reset has					
		er-on Reset occu	-	set in software	after a Power-	on Reset occurs	s)
bit 0		n-out Reset Stat					
		vn-out Reset has vn-out Reset occ					uro)
		m-out Reset oct	uneu (musi bi	e set in soltware		I-OUL Reset Occi	urs)
Noto 1:	It is recommon	ded that the \overline{POF}		or a Dowor on F	Posot has hoor	a datacted so th	at subsocue
		ets may be detec		ei a Fowei-Off	Coel nas Deel	ruelecieu, so in	at subsequel
	If the on-chip N BOR" for more	voltage regulator information.	is disabled, B	OR remains '0	' at all times. S	See Section 5.4	.1 "Detectin
3:	Brown-out Res	et is said to have	e occurred who	en BOR is '0' ar	nd POR is '1' (assuming that P	OR was set

REGISTER 5-1: RCON: RESET CONTROL REGISTER

'1' by software immediately after a Power-on Reset).

5.2 Master Clear (MCLR)

The MCLR pin provides a method for triggering a hard external Reset of the device. A Reset is generated by holding the pin low. PIC18 extended microcontroller devices have a noise filter in the MCLR Reset path which detects and ignores small pulses.

The $\overline{\text{MCLR}}$ pin is not driven low by any internal Resets, including the WDT.

5.3 Power-on Reset (POR)

A Power-on Reset condition is generated on-chip whenever VDD rises above a certain threshold. This allows the device to start in the initialized state when VDD is adequate for operation.

To take advantage of the POR circuitry, tie the $\overline{\text{MCLR}}$ pin through a resistor (1 k Ω to 10 k Ω) to VDD. This will eliminate external RC components usually needed to create a Power-on Reset delay. A minimum rise rate for VDD is specified (parameter D004). For a slow rise time, see Figure 5-2.

When the device starts normal operation (i.e., exits the Reset condition), device operating parameters (voltage, frequency, temperature, etc.) must be met to ensure operation. If these conditions are not met, the device must be held in Reset until the operating conditions are met.

Power-on Reset events are captured by the \overrightarrow{POR} bit (RCON<1>). The state of the bit is set to '0' whenever a Power-on Reset occurs; it does not change for any other Reset event. \overrightarrow{POR} is not reset to '1' by any hardware event. To capture multiple events, the user manually resets the bit to '1' in software following any Power-on Reset.

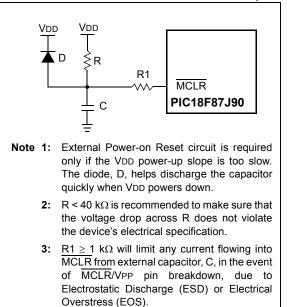
5.4 Brown-out Reset (BOR)

The PIC18F87J90 family of devices incorporates a simple BOR function when the internal regulator is enabled (ENVREG pin is tied to VDD). The voltage regulator will trigger a Brown-out Reset when output of the regulator to the device core approaches the voltage at which the device is unable to run at full speed. The BOR circuit also keeps the device in Reset as VDD rises, until the regulator's output level is sufficient for full-speed operation.

Once a BOR has occurred, the Power-up Timer will keep the chip in Reset for TPWRT (parameter 33). If VDD drops below the threshold for full-speed operation while the Power-up Timer is running, the chip will go back into a Brown-out Reset and the Power-up Timer will be initialized. Once VDD rises to the point where the regulator output is sufficient, the Power-up Timer will execute the additional time delay.

FIGURE 5-2:

EXTERNAL POWER-ON RESET CIRCUIT (FOR SLOW VDD POWER-UP)



5.4.1 DETECTING BOR

The BOR bit always resets to '0' on any Brown-out Reset or Power-on Reset event. This makes it difficult to determine if a Brown-out Reset event has occurred just by reading the state of BOR alone. A more reliable method is to simultaneously check the state of both POR and BOR. This assumes that the POR bit is reset to '1' in software immediately after any Power-on Reset event. If BOR is '0' while POR is '1', it can be reliably assumed that a Brown-out Reset event has occurred.

If the voltage regulator is disabled, Brown-out Reset functionality is disabled. In this case, the BOR bit cannot be used to determine a Brown-out Reset event. The BOR bit is still cleared by a Power-on Reset event.

5.5 Configuration Mismatch (CM)

The Configuration Mismatch (CM) Reset is designed to detect, and attempt to recover from, random, memory corrupting events. These include Electrostatic Discharge (ESD) events that can cause widespread, single bit changes throughout the device and result in catastrophic failure.

In PIC18FXXJ Flash devices, the device Configuration registers (located in the configuration memory space) are continuously monitored during operation by comparing their values to complimentary shadow registers. If a mismatch is detected between the two sets of registers, a CM Reset automatically occurs. These events are captured by the CM bit (RCON<5>). The state of the bit is set to '0' whenever a CM event occurs. The bit does not change for any other Reset event.

5.6 Power-up Timer (PWRT)

PIC18F87J90 family devices incorporate an on-chip Power-up Timer (PWRT) to help regulate the Power-on Reset process. The PWRT is always enabled. The main function is to ensure that the device voltage is stable before code is executed.

The Power-up Timer (PWRT) of the PIC18F87J90 family devices is an 11-bit counter which uses the INTRC source as the clock input. This yields an approximate time interval of 2048 x 32 μ s = 65.6 ms. While the PWRT is counting, the device is held in Reset.

The power-up time delay depends on the INTRC clock and will vary from chip to chip due to temperature and process variation. See DC parameter 33 for details.

5.6.1 TIME-OUT SEQUENCE

If enabled, the PWRT time-out is invoked after the POR pulse has cleared. The total time-out will vary based on the status of the PWRT. Figure 5-3, Figure 5-4, Figure 5-5 and Figure 5-6 all depict time-out sequences on power-up with the Power-up Timer enabled.

Since the time-outs occur from the POR pulse, if MCLR is kept low long enough, the PWRT will expire. Bringing MCLR high will begin execution immediately (Figure 5-5). This is useful for testing purposes or to synchronize more than one PIC18FXXXX device operating in parallel.

FIGURE 5-3: TIME-OUT SEQUENCE ON POWER-UP (MCLR TIED TO VDD, VDD RISE < TPWRT)

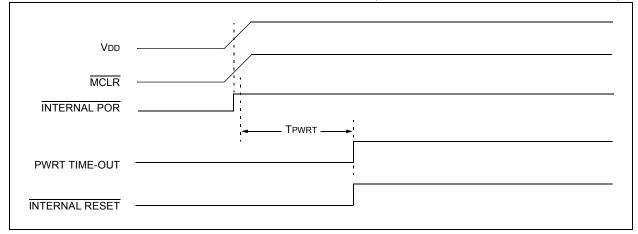
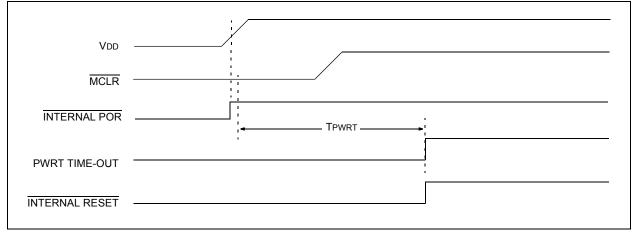


FIGURE 5-4: TIME-OUT SEQUENCE ON POWER-UP (MCLR NOT TIED TO VDD): CASE 1



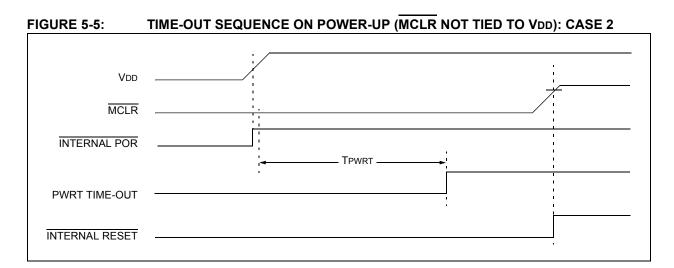
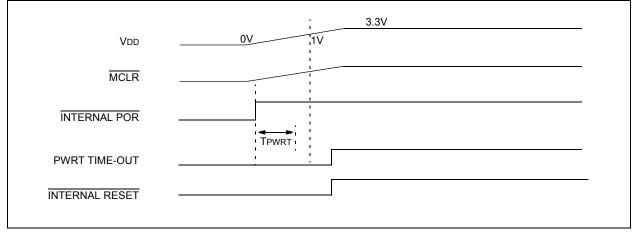


FIGURE 5-6: SLOW RISE TIME (MCLR TIED TO VDD, VDD RISE > TPWRT)



5.7 Reset State of Registers

Most registers are unaffected by a Reset. Their status is unknown on POR and unchanged by all other Resets. The other registers are forced to a "Reset state" depending on the type of Reset that occurred.

Most registers are not affected by a WDT wake-up, since this is viewed as the resumption of normal operation. Status bits from the RCON register, \overline{RI} , \overline{TO} , \overline{PD} , \overline{POR} and \overline{BOR} , are set or cleared differently in different Reset situations, as indicated in Table 5-1. These bits are used in software to determine the nature of the Reset.

Table 5-2 describes the Reset states for all of the Special Function Registers. These are categorized by Power-on and Brown-out Resets, Master Clear and WDT Resets, and WDT wake-ups.

TABLE 5-1:STATUS BITS, THEIR SIGNIFICANCE AND THE INITIALIZATION CONDITION FOR
RCON REGISTER

O a realitie re	Program		RC	STKPTR Register				
Condition	Counter ⁽¹⁾	RI	то	PD	POR	BOR	STKFUL	STKUNF
Power-on Reset	0000h	1	1	1	0	0	0	0
RESET Instruction	0000h	0	u	u	u	u	u	u
Brown-out Reset	0000h	1	1	1	u	0	u	u
MCLR during power-managed Run modes	0000h	u	1	u	u	u	u	u
MCLR during power-managed Idle modes and Sleep mode	0000h	u	1	0	u	u	u	u
WDT time-out during full power or power-managed Run modes	0000h	u	0	u	u	u	u	u
MCLR during full-power execution	0000h	u	u	u	u	u	u	u
Stack Full Reset (STVREN = 1)	0000h	u	u	u	u	u	1	u
Stack Underflow Reset (STVREN = 1)	0000h	u	u	u	u	u	u	1
Stack Underflow Error (not an actual Reset, STVREN = 0)	0000h	u	u	u	u	u	u	1
WDT time-out during power-managed Idle or Sleep modes	PC + 2	u	0	0	u	u	u	u
Interrupt exit from power-managed modes	PC + 2	u	u	0	u	u	u	u

Legend: u = unchanged

Note 1: When the wake-up is due to an interrupt and the GIEH or GIEL bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

PIC18F87J90 FAMILY

Register	Applicable Devices		Power-on Reset, Brown-out Reset	MCLR Resets WDT Reset RESET Instruction Stack Resets	Wake-up via WDT or Interrupt
TOSU	PIC18F6XJ90	PIC18F8XJ90	0 0000	0 0000	0 uuuu (1)
TOSH	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu ⁽¹⁾
TOSL	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu ⁽¹⁾
STKPTR	PIC18F6XJ90	PIC18F8XJ90	uu-0 0000	00-0 0000	uu-u uuuu ⁽¹⁾
PCLATU	PIC18F6XJ90	PIC18F8XJ90	0 0000	0 0000	u uuuu
PCLATH	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
PCL	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	PC + 2 ⁽²⁾
TBLPTRU	PIC18F6XJ90	PIC18F8XJ90	00 0000	00 0000	uu uuuu
TBLPTRH	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
TBLPTRL	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
TABLAT	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
PRODH	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu
PRODL	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu
INTCON	PIC18F6XJ90	PIC18F8XJ90	0000 000x	0000 000u	uuuu uuuu ⁽³⁾
INTCON2	PIC18F6XJ90	PIC18F8XJ90	1111 1111	1111 1111	uuuu uuuu ⁽³⁾
INTCON3	PIC18F6XJ90	PIC18F8XJ90	1100 0000	1100 0000	uuuu uuuu ⁽³⁾
INDF0	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A
POSTINC0	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A
POSTDEC0	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A
PREINC0	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A
PLUSW0	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A
FSR0H	PIC18F6XJ90	PIC18F8XJ90	xxxx	uuuu	uuuu
FSR0L	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu
WREG	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu
INDF1	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A
POSTINC1	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A
POSTDEC1	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A
PREINC1	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A
PLUSW1	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A

TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

Note 1: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

- 3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).
- 4: See Table 5-1 for Reset value for specific conditions.
- **5:** Bits 6 and 7 of PORTA, LATA and TRISA are enabled depending on the oscillator mode selected. When not enabled as PORTA pins, they are disabled and read as '0'.

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TABLE 5-2:	2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)						
Register	Applicable Devices		Power-on Reset, Brown-out Reset	MCLR Resets WDT Reset RESET Instruction Stack Resets	Wake-up via WDT or Interrupt		
FSR1H	PIC18F6XJ90	PIC18F8XJ90	xxxx	uuuu	uuuu		
FSR1L	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu		
BSR	PIC18F6XJ90	PIC18F8XJ90	0000	0000	uuuu		
INDF2	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A		
POSTINC2	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A		
POSTDEC2	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A		
PREINC2	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A		
PLUSW2	PIC18F6XJ90	PIC18F8XJ90	N/A	N/A	N/A		
FSR2H	PIC18F6XJ90	PIC18F8XJ90	xxxx	uuuu	uuuu		
FSR2L	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu		
STATUS	PIC18F6XJ90	PIC18F8XJ90	x xxxx	u uuuu	u uuuu		
TMR0H	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu		
TMR0L	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu		
TOCON	PIC18F6XJ90	PIC18F8XJ90	1111 1111	1111 1111	uuuu uuuu		
OSCCON	PIC18F6XJ90	PIC18F8XJ90	0110 q000	0110 q000	uuuu quuu		
LCDREG	PIC18F6XJ90	PIC18F8XJ90	-011 1100	-011 1000	-uuu uuuu		
WDTCON	PIC18F6XJ90	PIC18F8XJ90	00	00	uu		
RCON ⁽⁴⁾	PIC18F6XJ90	PIC18F8XJ90	0-11 11q0	0-0q qquu	u-uu qquu		
TMR1H	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu		
TMR1L	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu		
T1CON	PIC18F6XJ90	PIC18F8XJ90	0000 0000	u0uu uuuu	uuuu uuuu		
TMR2	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu		
PR2	PIC18F6XJ90	PIC18F8XJ90	1111 1111	1111 1111	1111 1111		
T2CON	PIC18F6XJ90	PIC18F8XJ90	-000 0000	-000 0000	-uuu uuuu		
SSPBUF	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu		
SSPADD	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu		
SSPSTAT	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu		
SSPCON1	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu		
SSPCON2	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu		

TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

Note 1: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

- 3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).
- 4: See Table 5-1 for Reset value for specific conditions.
- **5:** Bits 6 and 7 of PORTA, LATA and TRISA are enabled depending on the oscillator mode selected. When not enabled as PORTA pins, they are disabled and read as '0'.

Register	Applicable Devices		Power-on Reset, Brown-out Reset	MCLR Resets WDT Reset RESET Instruction Stack Resets	Wake-up via WDT or Interrupt
ADRESH	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu
ADRESL	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu
ADCON0	PIC18F6XJ90	PIC18F8XJ90	0-00 0000	0-00 0000	u-uu uuuu
ADCON1	PIC18F6XJ90	PIC18F8XJ90	0-00 0000	0-00 0000	u-uu uuuu
ADCON2	PIC18F6XJ90	PIC18F8XJ90	0-00 0000	0-00 0000	u-uu uuuu
LCDDATA4	PIC18F6XJ90	PIC18F8XJ90	x	u	u
LCDDATA4	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu
LCDDATA3	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu
LCDDATA2	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu
LCDDATA1	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu
LCDDATA0	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu
LCDSE5	PIC18F6XJ90	PIC18F8XJ90	0000 0000	uuuu uuuu	uuuu uuuu
LCDSE4	PIC18F6XJ90	PIC18F8XJ90	0	u	u
LCDSE4	PIC18F6XJ90	PIC18F8XJ90	0000 0000	uuuu uuuu	uuuu uuuu
LCDSE3	PIC18F6XJ90	PIC18F8XJ90	0000 0000	uuuu uuuu	uuuu uuuu
LCDSE2	PIC18F6XJ90	PIC18F8XJ90	0000 0000	uuuu uuuu	uuuu uuuu
LCDSE1	PIC18F6XJ90	PIC18F8XJ90	0000 0000	uuuu uuuu	uuuu uuuu
CVRCON	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
CMCON	PIC18F6XJ90	PIC18F8XJ90	0000 0111	0000 0111	uuuu uuuu
TMR3H	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu
TMR3L	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu
T3CON	PIC18F6XJ90	PIC18F8XJ90	0000 0000	uuuu uuuu	uuuu uuuu
SPBRG1	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
RCREG1	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
TXREG1	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
TXSTA1	PIC18F6XJ90	PIC18F8XJ90	0000 0010	0000 0010	uuuu uuuu
RCSTA1	PIC18F6XJ90	PIC18F8XJ90	0000 000x	0000 000x	սսսս սսսս
LCDPS	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
LCDSE0	PIC18F6XJ90	PIC18F8XJ90	0000 0000	uuuu uuuu	uuuu uuuu
LCDCON	PIC18F6XJ90	PIC18F8XJ90	000- 0000	000- 0000	uuu- uuuu
EECON2	PIC18F6XJ90	PIC18F8XJ90			
EECON1	PIC18F6XJ90	PIC18F8XJ90	0 x00-	0 u00-	0 u00-

TABLE 5-2:	INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

Note 1: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

- 2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).
- 3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).
- **4:** See Table 5-1 for Reset value for specific conditions.
- **5:** Bits 6 and 7 of PORTA, LATA and TRISA are enabled depending on the oscillator mode selected. When not enabled as PORTA pins, they are disabled and read as '0'.

PIC18F87J90 FAMILY

Register	Applicable Devices		Power-on Reset, Brown-out Reset	MCLR Resets WDT Reset RESET Instruction Stack Resets	D) Wake-up via WDT or Interrupt	
IPR3	PIC18F6XJ90	PIC18F8XJ90	-111 1111	-111 1111	-uuu 1111	
PIR3	PIC18F6XJ90	PIC18F8XJ90	-000 0000	-000 0000	-uuu 0000 (3)	
PIE3	PIC18F6XJ90	PIC18F8XJ90	-000 0000	-000 0000	-uuu 0000	
IPR2	PIC18F6XJ90	PIC18F8XJ90	11 111-	11 111-	uu uuu-	
PIR2	PIC18F6XJ90	PIC18F8XJ90	00 000-	00 000-	uu uuu- ⁽³⁾	
PIE2	PIC18F6XJ90	PIC18F8XJ90	00 000-	00 000-	uu uuu-	
IPR1	PIC18F6XJ90	PIC18F8XJ90	-111 1-11	-111 1-11	-uuu u-uu	
PIR1	PIC18F6XJ90	PIC18F8XJ90	-000 0-00	-000 0-00	-uuu u-uu ⁽³⁾	
PIE1	PIC18F6XJ90	PIC18F8XJ90	-000 0-00	-000 0-00	-uuu u-uu	
OSCTUNE	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu	
TRISJ	PIC18F6XJ90	PIC18F8XJ90	1111 1111	1111 1111	uuuu uuuu	
TRISH	PIC18F6XJ90	PIC18F8XJ90	1111 1111	1111 1111	uuuu uuuu	
TRISG	PIC18F6XJ90	PIC18F8XJ90	0001 1111	0001 1111	uuuu uuuu	
TRISF	PIC18F6XJ90	PIC18F8XJ90	1111 111-	1111 111-	uuuu uuu-	
TRISE	PIC18F6XJ90	PIC18F8XJ90	1111 1-11	1111 1-11	uuuu u-uu	
TRISD	PIC18F6XJ90	PIC18F8XJ90	1111 1111	1111 1111	uuuu uuuu	
TRISC	PIC18F6XJ90	PIC18F8XJ90	1111 1111	1111 1111	uuuu uuuu	
TRISB	PIC18F6XJ90	PIC18F8XJ90	1111 1111	1111 1111	uuuu uuuu	
TRISA ⁽⁵⁾	PIC18F6XJ90	PIC18F8XJ90	1111 1111 (5)	1111 1111 (5)	uuuu uuuu ⁽⁵⁾	
LATJ	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LATH	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LATG	PIC18F6XJ90	PIC18F8XJ90	00-x xxxx	00-u uuuu	uu-u uuuu	
LATF	PIC18F6XJ90	PIC18F8XJ90	xxxx xxx-	uuuu uuu-	uuuu uuu-	
LATE	PIC18F6XJ90	PIC18F8XJ90	xxxx x-xx	uuuu u-uu	uuuu u-uu	
LATD	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	
LATC	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	
LATB	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	
LATA ⁽⁵⁾	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx(5)	uuuu uuuu ⁽⁵⁾	uuuu uuuu ⁽⁵⁾	
PORTJ	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	
PORTH	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	
PORTG	PIC18F6XJ90	PIC18F8XJ90	000x xxxx	000u uuuu	000u uuuu	
PORTF	PIC18F6XJ90	PIC18F8XJ90	xxxx xxx-	uuuu uuu-	uuuu uuu-	

TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

Note 1: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

- 2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).
- 3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).
- 4: See Table 5-1 for Reset value for specific conditions.
- **5:** Bits 6 and 7 of PORTA, LATA and TRISA are enabled depending on the oscillator mode selected. When not enabled as PORTA pins, they are disabled and read as '0'.

TABLE 5-2:	INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)					
Register	Applicable Devices		Power-on Reset, Brown-out Reset	MCLR Resets WDT Reset RESET Instruction Stack Resets	Wake-up via WDT or Interrupt	
PORTE	PIC18F6XJ90	PIC18F8XJ90	xxxx x-xx	uuuu u-uu	uuuu u-uu	
PORTD	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	սսսս սսսս	
PORTC	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
PORTB	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
PORTA ⁽⁵⁾	PIC18F6XJ90	PIC18F8XJ90	xx0x 0000 ⁽⁵⁾	uu0u 0000 ⁽⁵⁾	uuuu uuuu ⁽⁵⁾	
SPBRGH1	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu	
BAUDCON1	PIC18F6XJ90	PIC18F8XJ90	0100 0-00	0100 0-00	uuuu u-uu	
LCDDATA23	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA22	PIC18F6XJ90	PIC18F8XJ90	x	u	u	
LCDDATA22	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA21	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	սսսս սսսս	
LCDDATA20	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA19	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA18	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	սսսս սսսս	
LCDDATA17	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA16	PIC18F6XJ90	PIC18F8XJ90	x	u	u	
LCDDATA16	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	սսսս սսսս	
LCDDATA15	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA14	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA13	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	սսսս սսսս	
LCDDATA12	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA11	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA10	PIC18F6XJ90	PIC18F8XJ90	x	u	u	
LCDDATA10	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA9	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA8	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu	
LCDDATA7	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	
LCDDATA6	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	սսսս սսսս	
LCDDATA5	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	
CCPR1H	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	
CCPR1L	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	սսսս սսսս	
CCP1CON	PIC18F6XJ90	PIC18F8XJ90	00 0000	00 0000	uu uuuu	
CCPR2H	PIC18F6XJ90	PIC18F8XJ90	XXXX XXXX	uuuu uuuu	uuuu uuuu	

TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

Note 1: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

- 4: See Table 5-1 for Reset value for specific conditions.
- **5:** Bits 6 and 7 of PORTA, LATA and TRISA are enabled depending on the oscillator mode selected. When not enabled as PORTA pins, they are disabled and read as '0'.

PIC18F87J90 FAMILY

TADLE 3-2.					
Register	Applicable Devices		Applicable Devices Power-on Reset, W Brown-out Reset RESE Sta		Wake-up via WDT or Interrupt
CCPR2L	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu
CCP2CON	PIC18F6XJ90	PIC18F8XJ90	00 0000	00 0000	uu uuuu
SPBRG2	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
RCREG2	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	սսսս սսսս
TXREG2	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
TXSTA2	PIC18F6XJ90	PIC18F8XJ90	0000 -010	0000 -010	uuuu -uuu
RCSTA2	PIC18F6XJ90	PIC18F8XJ90	0000 000x	0000 000x	սսսս սսսս
RTCCFG	PIC18F6XJ90	PIC18F8XJ90	0-00 0000	0-00 0000	u-uu uuuu
RTCCAL	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
RTCVALH	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	սսսս սսսս
RTCVALL	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	սսսս սսսս
ALRMCFG	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	uuuu uuuu
ALRMRPT	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	սսսս սսսս
ALRMVALH	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	սսսս սսսս
ALRMVALL	PIC18F6XJ90	PIC18F8XJ90	xxxx xxxx	uuuu uuuu	uuuu uuuu
CTMUCONH	PIC18F6XJ90	PIC18F8XJ90	0-00 0000	0-00 0000	u-uu uuuu
CTMUCONL	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	սսսս սսսս
CTMUICON	PIC18F6XJ90	PIC18F8XJ90	0000 0000	0000 0000	սսսս սսսս
PADCFG1	PIC18F6XJ90	PIC18F8XJ90	00-	00-	uu-

TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

 $\label{eq:logend:u} \begin{array}{ll} \mbox{u = unchanged, $x = unknown, $- = unimplemented bit, read as `0', $g = value depends on condition.} \\ Shaded cells indicate conditions do not apply for the designated device.} \end{array}$

Note 1: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

4: See Table 5-1 for Reset value for specific conditions.

5: Bits 6 and 7 of PORTA, LATA and TRISA are enabled depending on the oscillator mode selected. When not enabled as PORTA pins, they are disabled and read as '0'.

6.0 MEMORY ORGANIZATION

There are two types of memory in PIC18 Flash microcontroller devices:

- Program Memory
- Data RAM

As Harvard architecture devices, the data and program memories use separate busses; this allows for concurrent access of the two memory spaces.

Additional detailed information on the operation of the Flash program memory is provided in **Section 7.0 "Flash Program Memory"**.

6.1 Program Memory Organization

PIC18 microcontrollers implement a 21-bit program counter which is capable of addressing a 2-Mbyte program memory space. Accessing a location between the upper boundary of the physically implemented memory and the 2-Mbyte address will return all '0's (a NOP instruction).

The entire PIC18F87J90 family offers a range of on-chip Flash program memory sizes, from 64 Kbytes (up to 16,384 single-word instructions) to 128 Kbytes (65,536 single-word instructions). The program memory maps for individual family members are shown in Figure 6-1.

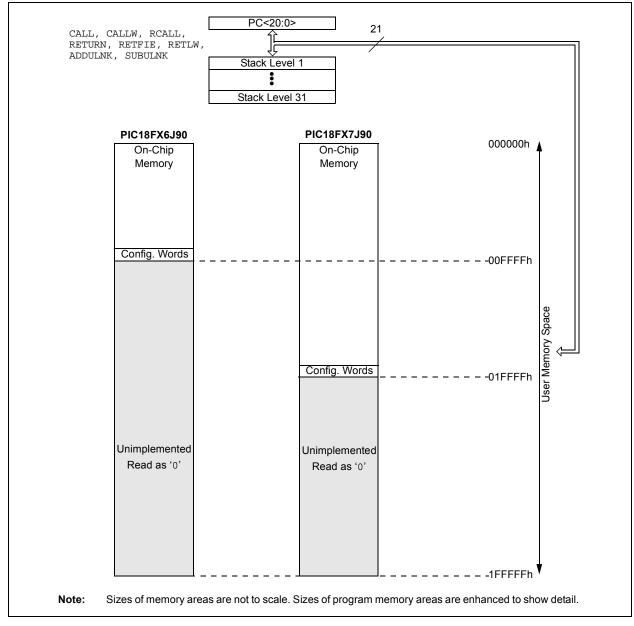


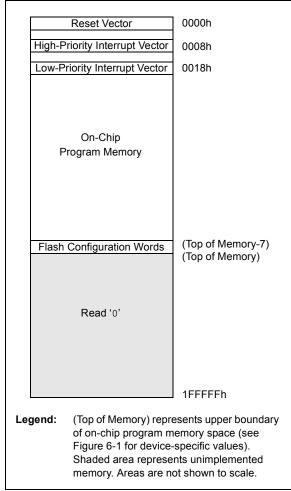
FIGURE 6-1: MEMORY MAPS FOR PIC18F87J90 FAMILY DEVICES

6.1.1 HARD MEMORY VECTORS

All PIC18 devices have a total of three hard-coded return vectors in their program memory space. The Reset vector address is the default value to which the program counter returns on all device Resets; it is located at 0000h.

PIC18 devices also have two interrupt vector addresses for the handling of high-priority and low-priority interrupts. The high-priority interrupt vector is located at 0008h and the low-priority interrupt vector is at 0018h. Their locations in relation to the program memory map are shown in Figure 6-2.

FIGURE 6-2: HARD VECTOR AND CONFIGURATION WORD LOCATIONS FOR PIC18F87J90 FAMILY FAMILY DEVICES



6.1.2 FLASH CONFIGURATION WORDS

Because PIC18F87J90 family devices do not have persistent configuration memory, the top four words of on-chip program memory are reserved for configuration information. On Reset, the configuration information is copied into the Configuration registers.

The Configuration Words are stored in their program memory location in numerical order, starting with the lower byte of CONFIG1 at the lowest address and ending with the upper byte of CONFIG4. For these devices, only Configuration Words, CONFIG1 through CONFIG3, are used; CONFIG4 is reserved. The actual addresses of the Flash Configuration Word for devices in the PIC18F87J90 family are shown in Table 6-1. Their location in the memory map is shown with the other memory vectors in Figure 6-2.

Additional details on the device Configuration Words are provided in **Section 25.1** "Configuration Bits".

TABLE 6-1:	FLASH CONFIGURATION
	WORD FOR PIC18F87J90
	FAMILY DEVICES

Device	Program Memory (Kbytes)	Configuration Word Addresses	
PIC18F66J90	64	FFF8h to FFFFh	
PIC18F86J90	04		
PIC18F67J90	128	1FFF8h to 1FFFFh	
PIC18F87J90	120		

6.1.3 PROGRAM COUNTER

The Program Counter (PC) specifies the address of the instruction to fetch for execution. The PC is 21 bits wide and is contained in three separate 8-bit registers. The low byte, known as the PCL register, is both readable and writable. The high byte, or PCH register, contains the PC<15:8> bits; it is not directly readable or writable. Updates to the PCH register are performed through the PCLATH register. The upper byte is called PCU. This register contains the PC<20:16> bits; it is also not directly readable or writable. Updates to the PCH register. Updates to the PCU register are performed through the PCLATH register or writable. Updates to the PCU register are performed through the PCU register are performed through the PCU register are performed through the PCLATU register.

The contents of PCLATH and PCLATU are transferred to the program counter by any operation that writes PCL. Similarly, the upper two bytes of the program counter are transferred to PCLATH and PCLATU by an operation that reads PCL. This is useful for computed offsets to the PC (see **Section 6.1.6.1 "Computed GOTO"**).

The PC addresses bytes in the program memory. To prevent the PC from becoming misaligned with word instructions, the Least Significant bit of PCL is fixed to a value of '0'. The PC increments by 2 to address sequential instructions in the program memory.

The CALL, RCALL, GOTO and program branch instructions write to the program counter directly. For these instructions, the contents of PCLATH and PCLATU are not transferred to the program counter.

6.1.4 RETURN ADDRESS STACK

The return address stack allows any combination of up to 31 program calls and interrupts to occur. The PC is pushed onto the stack when a CALL or RCALL instruction is executed, or an interrupt is Acknowledged. The PC value is pulled off the stack on a RETURN, RETLW or a RETFIE instruction (and on ADDULNK and SUBULNK instructions if the extended instruction set is enabled). PCLATU and PCLATH are not affected by any of the RETURN or CALL instructions. The stack operates as a 31-word by 21-bit RAM and a 5-bit Stack Pointer, STKPTR. The stack space is not part of either program or data space. The Stack Pointer is readable and writable, and the address on the top of the stack is readable and writable through the Top-of-Stack Special Function Registers. Data can also be pushed to, or popped from, the stack, using these registers.

A CALL type instruction causes a push onto the stack. The Stack Pointer is first incremented and the location pointed to by the Stack Pointer is written with the contents of the PC (already pointing to the instruction following the CALL). A RETURN type instruction causes a pop from the stack. The contents of the location pointed to by the STKPTR are transferred to the PC and then the Stack Pointer is decremented.

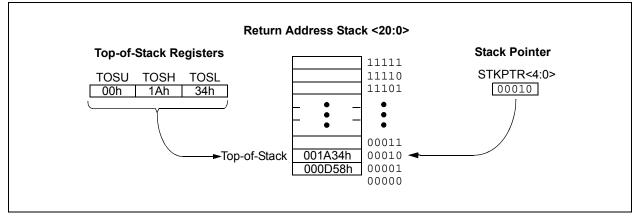
The Stack Pointer is initialized to '00000' after all Resets. There is no RAM associated with the location corresponding to a Stack Pointer value of '00000'; this is only a Reset value. Status bits indicate if the stack is full, has overflowed or has underflowed.

6.1.4.1 Top-of-Stack Access

Only the top of the return address stack (TOS) is readable and writable. A set of three registers, TOSU:TOSH:TOSL, holds the contents of the stack location pointed to by the STKPTR register (Figure 6-3). This allows users to implement a software stack, if necessary. After a CALL, RCALL or interrupt (and ADDULNK and SUBULNK instructions if the extended instruction set is enabled), the software can read the pushed value by reading the TOSU:TOSH:TOSL registers. These values can be placed on a user-defined software stack. At return time, the software can return these values to TOSU:TOSH:TOSL and do a return.

The user must disable the global interrupt enable bits while accessing the stack to prevent inadvertent stack corruption.





6.1.4.2 Return Stack Pointer (STKPTR)

The STKPTR register (Register 6-1) contains the Stack Pointer value, the STKFUL (Stack Full) status bit and the STKUNF (Stack Underflow) status bit. The value of the Stack Pointer can be 0 through 31. The Stack Pointer increments before values are pushed onto the stack and decrements after values are popped off of the stack. On Reset, the Stack Pointer value will be zero. The user may read and write the Stack Pointer value. This feature can be used by a Real-Time Operating System (RTOS) for return stack maintenance.

After the PC is pushed onto the stack 31 times (without popping any values off the stack), the STKFUL bit is set. The STKFUL bit is cleared by software or by a POR.

The action that takes place when the stack becomes full depends on the state of the STVREN (Stack Overflow Reset Enable) Configuration bit. (Refer to **Section 25.1 "Configuration Bits"** for a description of the device Configuration bits.) If STVREN is set (default), the 31st push will push the (PC + 2) value onto the stack, set the STKFUL bit and reset the device. The STKFUL bit will remain set and the Stack Pointer will be set to zero.

If STVREN is cleared, the STKFUL bit will be set on the 31st push and the Stack Pointer will increment to 31. Any additional pushes will not overwrite the 31st push and the STKPTR will remain at 31.

When the stack has been popped enough times to unload the stack, the next pop will return a value of zero to the PC and set the STKUNF bit, while the Stack Pointer remains at zero. The STKUNF bit will remain set until cleared by software or until a POR occurs.

Note:	Returning a value of zero to the PC on an underflow has the effect of vectoring the program to the Reset vector, where the stack conditions can be verified and appropriate actions can be taken. This is not the same as a Reset as the contents of
	the SFRs are not affected.

6.1.4.3 PUSH and POP Instructions

Since the Top-of-Stack is readable and writable, the ability to push values onto the stack and pull values off the stack, without disturbing normal program execution, is a desirable feature. The PIC18 instruction set includes two instructions, PUSH and POP, that permit the TOS to be manipulated under software control. TOSU, TOSH and TOSL can be modified to place data or a return address on the stack.

The PUSH instruction places the current PC value onto the stack. This increments the Stack Pointer and loads the current PC value onto the stack.

The POP instruction discards the current TOS by decrementing the Stack Pointer. The previous value pushed onto the stack then becomes the TOS value.

R/C-0 R/C-0 U-0 R/W-0 R/W-0 R/W-0 R/W-0 R/W-0 STKFUL⁽¹⁾ STKUNF⁽¹⁾ SP4 SP3 SP2 SP1 SP0 bit 7 bit 0 Legend: C = Clearable bit R = Readable bit W = Writable bit U = Unimplemented bit, read as '0' '0' = Bit is cleared -n = Value at POR '1' = Bit is set x = Bit is unknown STKFUL: Stack Full Flag bit⁽¹⁾ bit 7 1 = Stack became full or overflowed 0 = Stack has not become full or overflowed STKUNF: Stack Underflow Flag bit⁽¹⁾ bit 6 1 = Stack underflow occurred

REGISTER 6-1: STKPTR: STACK POINTER REGISTER

bit 5 Unimplemented: Read as '0'

bit 4-0 **SP<4:0>:** Stack Pointer Location bits

Note 1: Bit 7 and bit 6 are cleared by user software or by a POR.

0 = Stack underflow did not occur

6.1.4.4 Stack Full and Underflow Resets

Device Resets on stack overflow and stack underflow conditions are enabled by setting the STVREN bit in Configuration Register 1L. When STVREN is set, a full or underflow condition will set the appropriate STKFUL or STKUNF bit and then cause a device Reset. When STVREN is cleared, a full or underflow condition will set the appropriate STKFUL or STKUNF bit, but not cause a device Reset. The STKFUL or STKUNF bits are cleared by the user software or a Power-on Reset.

6.1.5 FAST REGISTER STACK

A Fast Register Stack is provided for the STATUS, WREG and BSR registers to provide a "fast return" option for interrupts. This stack is only one level deep and is neither readable nor writable. It is loaded with the current value of the corresponding register when the processor vectors for an interrupt. All interrupt sources will push values into the Stack registers. The values in the registers are then loaded back into the working registers if the RETFIE, FAST instruction is used to return from the interrupt.

If both low and high-priority interrupts are enabled, the Stack registers cannot be used reliably to return from low-priority interrupts. If a high-priority interrupt occurs while servicing a low-priority interrupt, the Stack register values stored by the low-priority interrupt will be overwritten. In these cases, users must save the key registers in software during a low-priority interrupt.

If interrupt priority is not used, all interrupts may use the Fast Register Stack for returns from interrupt. If no interrupts are used, the Fast Register Stack can be used to restore the STATUS, WREG and BSR registers at the end of a subroutine call. To use the Fast Register Stack for a subroutine call, a CALL label, FAST instruction must be executed to save the STATUS, WREG and BSR registers to the Fast Register Stack. A RETURN, FAST instruction is then executed to restore these registers from the Fast Register Stack.

Example 6-1 shows a source code example that uses the Fast Register Stack during a subroutine call and return.

EXAMPLE 6-1: FAST REGISTER STACK CODE EXAMPLE

CALL	SUB1, FAST •	;STATUS, WREG, BSR ;SAVED IN FAST REGISTER ;STACK
SUB1	• • RETURN FAST	;RESTORE VALUES SAVED ;IN FAST REGISTER STACK

6.1.6 LOOK-UP TABLES IN PROGRAM MEMORY

There may be programming situations that require the creation of data structures, or look-up tables, in program memory. For PIC18 devices, look-up tables can be implemented in two ways:

- Computed GOTO
- Table Reads

6.1.6.1 Computed GOTO

A computed GOTO is accomplished by adding an offset to the program counter. An example is shown in Example 6-2.

A look-up table can be formed with an ADDWF PCL instruction and a group of RETLW nn instructions. The W register is loaded with an offset into the table before executing a call to that table. The first instruction of the called routine is the ADDWF PCL instruction. The next instruction executed will be one of the RETLW nn instructions that returns the value 'nn' to the calling function.

The offset value (in WREG) specifies the number of bytes that the program counter should advance and should be multiples of 2 (LSb = 0).

In this method, only one data byte may be stored in each instruction location and room on the return address stack is required.

EXAMPLE 6-2: COMPUTED GOTO USING AN OFFSET VALUE

		/
	MOVF	OFFSET, W
	CALL	TABLE
ORG	nn00h	
TABLE	ADDWF	PCL
	RETLW	nnh
	RETLW	nnh
	RETLW	nnh

6.1.6.2 Table Reads

A better method of storing data in program memory allows two bytes of data to be stored in each instruction location.

Look-up table data may be stored two bytes per program word while programming. The Table Pointer (TBLPTR) specifies the byte address and the Table Latch (TABLAT) contains the data that is read from the program memory. Data is transferred from program memory, one byte at a time.

Table read operation is discussed further in **Section 7.1 "Table Reads and Table Writes**".

6.2 PIC18 Instruction Cycle

6.2.1 CLOCKING SCHEME

The microcontroller clock input, whether from an internal or external source, is internally divided by four to generate four non-overlapping quadrature clocks (Q1, Q2, Q3 and Q4). Internally, the program counter is incremented on every Q1; the instruction is fetched from the program memory and latched into the Instruction Register (IR) during Q4. The instruction is decoded and executed during the following Q1 through Q4. The clocks and instruction execution flow are shown in Figure 6-4.

6.2.2 INSTRUCTION FLOW/PIPELINING

An "Instruction Cycle" consists of four Q cycles, Q1 through Q4. The instruction fetch and execute are pipelined in such a manner that a fetch takes one instruction cycle, while the decode and execute take another instruction cycle. However, due to the pipelining, each instruction effectively executes in one cycle. If an instruction causes the program counter to change (e.g., GOTO), then two cycles are required to complete the instruction (Example 6-3).

A fetch cycle begins with the Program Counter (PC) incrementing in Q1.

In the execution cycle, the fetched instruction is latched into the Instruction Register (IR) in cycle, Q1. This instruction is then decoded and executed during the Q2, Q3 and Q4 cycles. Data memory is read during Q2 (operand read) and written during Q4 (destination write).

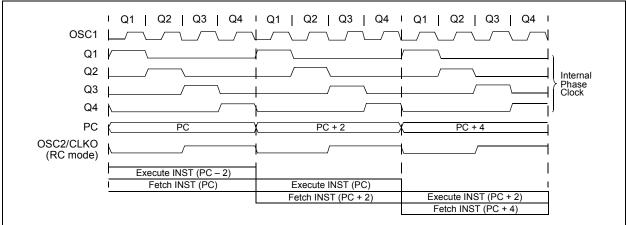
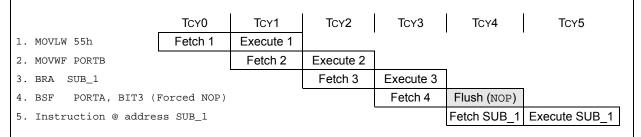


FIGURE 6-4: CLOCK/INSTRUCTION CYCLE

EXAMPLE 6-3: INSTRUCTION PIPELINE FLOW



All instructions are single cycle, except for any program branches. These take two cycles since the fetch instruction is "flushed" from the pipeline while the new instruction is being fetched and then executed.

6.2.3 INSTRUCTIONS IN PROGRAM MEMORY

The program memory is addressed in bytes. Instructions are stored as two bytes or four bytes in program memory. The Least Significant Byte (LSB) of an instruction word is always stored in a program memory location with an even address (LSB = 0). To maintain alignment with instruction boundaries, the PC increments in steps of 2 and the LSB will always read '0' (see Section 6.1.3 "Program Counter").

Figure 6-5 shows an example of how instruction words are stored in the program memory.

The CALL and GOTO instructions have the absolute program memory address embedded into the instruction. Since instructions are always stored on word boundaries, the data contained in the instruction is a word address. The word address is written to PC<20:1> which accesses the desired byte address in program memory. Instruction #2 in Figure 6-5 shows how the instruction, GOTO 0006h, is encoded in the program memory. Program branch instructions, which encode a relative address offset, operate in the same manner. The offset value stored in a branch instruction represents the number of single-word instructions that the PC will be offset by. Section 26.0 "Instruction Set Summary" provides further details of the instruction set.

FIGURE 6-5: INSTRUCTIONS IN PROGRAM MEMORY	FIGURE 6-5:	INSTRUCTIONS IN PROGRAM MEMORY
--	-------------	--------------------------------

			LSB = 1	LSB = 0	Word Address \downarrow
	Program Memory				000000h
	Byte Locat	ions \rightarrow			000002h
					000004h
					000006h
Instruction 1:	MOVLW	055h	0Fh	55h	000008h
Instruction 2:	GOTO	0006h	EFh	03h	00000Ah
			F0h	00h	00000Ch
Instruction 3:	MOVFF	123h, 456h	C1h	23h	00000Eh
			F4h	56h	000010h
					000012h
					000014h

6.2.4 TWO-WORD INSTRUCTIONS

The standard PIC18 instruction set has four, two-word instructions: CALL, MOVFF, GOTO and LSFR. In all cases, the second word of the instructions always has '1111' as its four Most Significant bits (MSb); the other 12 bits are literal data, usually a data memory address.

The use of '1111' in the 4 MSbs of an instruction specifies a special form of NOP. If the instruction is executed in proper sequence, immediately after the first word, the data in the second word is accessed and

used by the instruction sequence. If the first word is skipped for some reason and the second word is executed by itself, a NOP is executed instead. This is necessary for cases when the two-word instruction is preceded by a conditional instruction that changes the PC. Example 6-4 shows how this works.

Note: See Section 6.5 "Program Memory and the Extended Instruction Set" for information on two-word instructions in the extended instruction set.

CASE 1:					
Object Code	bject Code Source Code				
0110 0110 0000 0000	TSTFSZ REG1	; is RAM location 0?			
1100 0001 0010 0011	MOVFF REG1, REG2	; No, skip this word			
1111 0100 0101 0110		; Execute this word as a NOP			
0010 0100 0000 0000	ADDWF REG3	; continue code			
CASE 2:					
Object Code	Source Code				
0110 0110 0000 0000	TSTFSZ REG1	; is RAM location 0?			
1100 0001 0010 0011	MOVFF REG1, REG2	; Yes, execute this word			
1111 0100 0101 0110		; 2nd word of instruction			
0010 0100 0000 0000	ADDWF REG3	; continue code			

6.3 Data Memory Organization

Note:	The operation of some aspects of data
	memory are changed when the PIC18
	extended instruction set is enabled. See
	Section 6.6 "Data Memory and the
	Extended Instruction Set" for more
	information.

The data memory in PIC18 devices is implemented as static RAM. Each register in the data memory has a 12-bit address, allowing up to 4,096 bytes of data memory. The memory space is divided into as many as 16 banks that contain 256 bytes each. PIC18FX6J90 and PIC18FX7J90 devices implement all 16 complete banks, for a total of 4 Kbytes. Figure 6-6 and Figure 6-7 show the data memory organization for the devices.

The data memory contains Special Function Registers (SFRs) and General Purpose Registers (GPRs). The SFRs are used for control and status of the controller and peripheral functions, while GPRs are used for data storage and scratchpad operations in the user's application. Any read of an unimplemented location will read as '0's.

The instruction set and architecture allow operations across all banks. The entire data memory may be accessed by Direct, Indirect or Indexed Addressing modes. Addressing modes are discussed later in this section.

To ensure that commonly used registers (select SFRs and select GPRs) can be accessed in a single cycle, PIC18 devices implement an Access Bank. This is a 256-byte memory space that provides fast access to select SFRs, and the lower portion of GPR Bank 0, without using the BSR. **Section 6.3.2 "Access Bank"** provides a detailed description of the Access RAM.

6.3.1 BANK SELECT REGISTER

Large areas of data memory require an efficient addressing scheme to make rapid access to any address possible. Ideally, this means that an entire address does not need to be provided for each read or write operation. For PIC18 devices, this is accomplished with a RAM banking scheme. This divides the memory space into 16 contiguous banks of 256 bytes. Depending on the instruction, each location can be addressed directly by its full 12-bit address, or an 8-bit low-order address and a 4-bit Bank Pointer.

Most instructions in the PIC18 instruction set make use of the Bank Pointer, known as the Bank Select Register (BSR). This SFR holds the 4 Most Significant bits of a location's address; the instruction itself includes the 8 Least Significant bits. Only the four lower bits of the BSR are implemented (BSR<3:0>). The upper four bits are unused; they will always read '0' and cannot be written to. The BSR can be loaded directly by using the MOVLB instruction.

The value of the BSR indicates the bank in data memory. The 8 bits in the instruction show the location in the bank and can be thought of as an offset from the bank's lower boundary. The relationship between the BSR's value and the bank division in data memory is shown in Figure 6-7.

Since up to 16 registers may share the same low-order address, the user must always be careful to ensure that the proper bank is selected before performing a data read or write. For example, writing what should be program data to an 8-bit address of F9h, while the BSR is 0Fh, will end up resetting the program counter.

While any bank can be selected, only those banks that are actually implemented can be read or written to. Writes to unimplemented banks are ignored, while reads from unimplemented banks will return '0's. Even so, the STATUS register will still be affected as if the operation was successful. The data memory map in Figure 6-6 indicates which banks are implemented.

In the core PIC18 instruction set, only the MOVFF instruction fully specifies the 12-bit address of the source and target registers. This instruction ignores the BSR completely when it executes. All other instructions include only the low-order address as an operand and must use either the BSR or the Access Bank to locate their target registers.

FIGURE 6-6: DATA MEMORY MAP FOR PIC18FX6J90 AND PIC18FX7J90 DEVICES When a = 0: BSR<3:0> **Data Memory Map** The BSR is ignored and the Access Bank is used. 00h 000h Access RAM The first 96 bytes are general 05Fh = 0000 Bank 0 purpose RAM (from Bank 0). 060h GPR FFh 0FFh The second 160 bytes are 00h 100h Special Function Registers = 0001 (from Bank 15). Bank 1 GPR FFh 1FFh 200h 00h When a = 1: = 0010 Bank 2 GPR The BSR specifies the bank FFh 2FFh used by the instruction. 300h 00h = 0011 Bank 3 GPR FFh 3FFh 00h 400h = 0100 Bank 4 GPR FFh 4FFh 00h 500h = 0101 GPR Bank 5 FFh 5FFh 00h 600h = 0110 GPR Bank 6 FFh 6FFh 700h 00h Access Bank = 0111 GPR Bank 7 00h FFh 7FFh Access RAM Low 5Fh 00h 800h = 1000 60h Access RAM High GPR Bank 8 (SFRs) FFh FFh 8FFh 00h 900h = 1001 Bank 9 GPR FFh 9FFh 00h A00h = 1010 Bank 10 GPR AFFh FFh B00h 00h = 1011 Bank 11 GPR BFFh FFh 00h C00h = 1100 Bank 12 GPR FFh CFFh D00h 00h = 1101 Bank 13 GPR DFFh FFh 00h E00h = 1110 Bank 14 GPR EFFh FFh F00h 00h = 1111 GPR(1) F5Fh Bank 15 F60h/ SFR FFh FFFh **Note 1:** Addresses, F54h through F5Fh, are also used by SFRs, but are not part of the Access RAM. Users must always use the complete address, or load the proper BSR value, to access these registers.

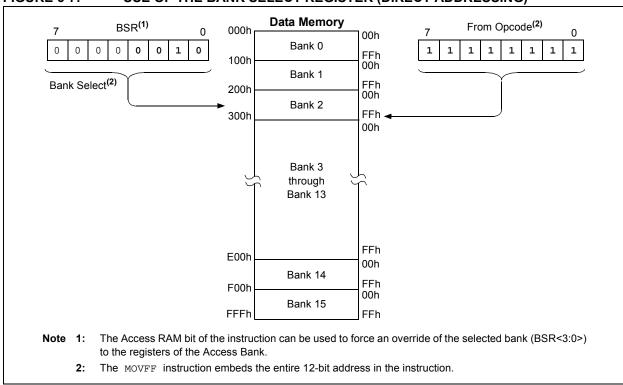


FIGURE 6-7: USE OF THE BANK SELECT REGISTER (DIRECT ADDRESSING)

6.3.2 ACCESS BANK

While the use of the BSR with an embedded 8-bit address allows users to address the entire range of data memory, it also means that the user must always ensure that the correct bank is selected. Otherwise, data may be read from, or written to, the wrong location. This can be disastrous if a GPR is the intended target of an operation, but an SFR is written to instead. Verifying and/or changing the BSR for each read or write to data memory can become very inefficient.

To streamline access for the most commonly used data memory locations, the data memory is configured with an Access Bank, which allows users to access a mapped block of memory without specifying a BSR. The Access Bank consists of the first 96 bytes of memory (00h-5Fh) in Bank 0 and the last 160 bytes of memory (60h-FFh) in Bank 15. The lower half is known as the "Access RAM" and is composed of GPRs. The upper half is where the device's SFRs are mapped. These two areas are mapped contiguously in the Access Bank and can be addressed in a linear fashion by an 8-bit address (Figure 6-6).

The Access Bank is used by core PIC18 instructions that include the Access RAM bit (the 'a' parameter in the instruction). When 'a' is equal to '1', the instruction uses the BSR and the 8-bit address included in the opcode for the data memory address. When 'a' is '0',

however, the instruction is forced to use the Access Bank address map; the current value of the BSR is ignored entirely.

Using this "forced" addressing allows the instruction to operate on a data address in a single cycle without updating the BSR first. For 8-bit addresses of 60h and above, this means that users can evaluate and operate on SFRs more efficiently. The Access RAM below 60h is a good place for data values that the user might need to access rapidly, such as immediate computational results or common program variables. Access RAM also allows for faster and more code efficient context saving and switching of variables.

The mapping of the Access Bank is slightly different when the extended instruction set is enabled (XINST Configuration bit = 1). This is discussed in more detail in Section 6.6.3 "Mapping the Access Bank in Indexed Literal Offset Mode".

6.3.3 GENERAL PURPOSE REGISTER FILE

PIC18 devices may have banked memory in the GPR area. This is data RAM which is available for use by all instructions. GPRs start at the bottom of Bank 0 (address 000h) and grow upwards towards the bottom of the SFR area. GPRs are not initialized by a Power-on Reset and are unchanged on all other Resets.

6.3.4 SPECIAL FUNCTION REGISTERS

The Special Function Registers (SFRs) are registers used by the CPU and peripheral modules for controlling the desired operation of the device. These registers are implemented as static RAM. SFRs start at the top of data memory (FFFh) and extend downward to occupy more than the top half of Bank 15 (F60h to FFFh). A list of these registers is given in Table 6-2 and Table 6-3. The SFRs can be classified into two sets: those associated with the "core" device functionality (ALU, Resets and interrupts) and those related to the peripheral functions. The Reset and Interrupt registers are described in their respective chapters, while the ALU's STATUS register is described later in this section. Registers related to the operation of the peripheral features are described in the chapter for that peripheral.

The SFRs are typically distributed among the peripherals whose functions they control. Unused SFR locations are unimplemented and read as '0's.

Addr.	Name	Addr.	Name	Addr.	Name	Addr.	Name	Addr.	Name	Addr.	Name
FFFh	TOSU	FDFh	INDF2 ⁽¹⁾	FBFh	LCDDATA4 ⁽³⁾	F9Fh	IPR1	F7Fh	SPBRGH1	F5Fh	RTCCFG
FFEh	TOSH	FDEh	POSTINC2 ⁽¹⁾	FBEh	LCDDATA3	F9Eh	PIR1	F7Eh	BAUDCON1	F5Eh	RTCCAL
FFDh	TOSL	FDDh	POSTDEC2 ⁽¹⁾	FBDh	LCDDATA2	F9Dh	PIE1	F7Dh	LCDDATA23 ⁽³⁾	F5Dh	RTCVALH
FFCh	STKPTR	FDCh	PREINC2 ⁽¹⁾	FBCh	LCDDATA1	F9Ch	(2)	F7Ch	LCDDATA22 ⁽³⁾	F5Ch	RTCVALL
FFBh	PCLATU	FDBh	PLUSW2 ⁽¹⁾	FBBh	LCDDATA0	F9Bh	OSCTUNE	F7Bh	LCDDATA21	F5Bh	ALRMCFG
FFAh	PCLATH	FDAh	FSR2H	FBAh	LCDSE5 ⁽³⁾	F9Ah	TRISJ ⁽³⁾	F7Ah	LCDDATA20	F5Ah	ALRMRPT
FF9h	PCL	FD9h	FSR2L	FB9h	LCDSE4 ⁽³⁾	F99h	TRISH ⁽³⁾	F79h	LCDDATA19	F59h	ALRMVALH
FF8h	TBLPTRU	FD8h	STATUS	FB8h	LCDSE3	F98h	TRISG	F78h	LCDDATA18	F58h	ALRMVALL
FF7h	TBLPTRH	FD7h	TMR0H	FB7h	LCDSE2	F97h	TRISF	F77h	LCDDATA17 ⁽³⁾	F57h	CTMUCONH
FF6h	TBLPTRL	FD6h	TMR0L	FB6h	LCDSE1	F96h	TRISE	F76h	LCDDATA16 ⁽³⁾	F56h	CTMUCONL
FF5h	TABLAT	FD5h	T0CON	FB5h	CVRCON	F95h	TRISD	F75h	LCDDATA15	F55h	CTMUICON
FF4h	PRODH	FD4h	(2)	FB4h	CMCON	F94h	TRISC	F74h	LCDDATA14	F54h	PADCFG1
FF3h	PRODL	FD3h	OSCCON	FB3h	TMR3H	F93h	TRISB	F73h	LCDDATA13		
FF2h	INTCON	FD2h	LCDREG	FB2h	TMR3L	F92h	TRISA	F72h	LCDDATA12		
FF1h	INTCON2	FD1h	WDTCON	FB1h	T3CON	F91h	LATJ ⁽³⁾	F71h	LCDDATA11 ⁽³⁾		
FF0h	INTCON3	FD0h	RCON	FB0h	(2)	F90h	LATH ⁽³⁾	F70h	LCDDATA10 ⁽³⁾		
FEFh	INDF0 ⁽¹⁾	FCFh	TMR1H	FAFh	SPBRG1	F8Fh	LATG	F6Fh	LCDDATA9		
FEEh	POSTINC0 ⁽¹⁾	FCEh	TMR1L	FAEh	RCREG1	F8Eh	LATF	F6Eh	LCDDATA8		
FEDh	POSTDEC0 ⁽¹⁾	FCDh	T1CON	FADh	TXREG1	F8Dh	LATE	F6Dh	LCDDATA7		
FECh	PREINC0 ⁽¹⁾	FCCh	TMR2	FACh	TXSTA1	F8Ch	LATD	F6Ch	LCDDATA6		
FEBh	PLUSW0 ⁽¹⁾	FCBh	PR2	FABh	RCSTA1	F8Bh	LATC	F6Bh	LCDDATA5 ⁽³⁾		
FEAh	FSR0H	FCAh	T2CON	FAAh	LCDPS	F8Ah	LATB	F6Ah	CCPR1H		
FE9h	FSR0L	FC9h	SSPBUF	FA9h	LCDSE0	F89h	LATA	F69h	CCPR1L		
FE8h	WREG	FC8h	SSPADD	FA8h	LCDCON	F88h	PORTJ ⁽³⁾	F68h	CCP1CON		
FE7h	INDF1 ⁽¹⁾	FC7h	SSPSTAT	FA7h	EECON2	F87h	PORTH ⁽³⁾	F67h	CCPR2H		
FE6h	POSTINC1 ⁽¹⁾	FC6h	SSPCON1	FA6h	EECON1	F86h	PORTG	F66h	CCPR2L		
FE5h	POSTDEC1 ⁽¹⁾	FC5h	SSPCON2	FA5h	IPR3	F85h	PORTF	F65h	CCP2CON		
FE4h		FC4h	ADRESH	FA4h	PIR3	F84h	PORTE	F64h	SPBRG2		
FE3h	PLUSW1 ⁽¹⁾	FC3h	ADRESL	FA3h	PIE3	F83h	PORTD	F63h	RCREG2		
FE2h	FSR1H	FC2h	ADCON0	FA2h	IPR2	F82h	PORTC	F62h	TXREG2		
FE1h	FSR1L	FC1h	ADCON1	FA1h	PIR2	F81h	PORTB	F61h	TXSTA2		
FE0h	BSR	FC0h	ADCON2	FA0h	PIE2	F80h	PORTA	F60h	RCSTA2		

TABLE 6-2: SPECIAL FUNCTION REGISTER MAP FOR PIC18F87J90 FAMILY DEVICES

Note 1: This is not a physical register.

2: Unimplemented registers are read as '0'.

3: This register is not available on PIC18F6XJ90 devices.

TABLE 6-3:	PIC18F87J90 FAMILY REGISTER FILE SUMMARY

File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Details on page	
TOSU	_	_	_	Top-of-Stack	Upper Byte (T	OS<20:16>)			0 0000	59, 67	
TOSH	Top-of-Stack	High Byte (TO	S<15:8>)						0000 0000	59, 67	
TOSL	Top-of-Stack	Low Byte (TO	S<7:0>)						0000 0000	59, 67	
STKPTR	STKFUL	STKUNF	_	Return Stack	Pointer				uu-0 0000	59, 68	
PCLATU	_	_	bit 21 ⁽¹⁾	Holding Regi	ster for PC<20):16>			0 0000	59, 67	
PCLATH	Holding Regis	ster for PC<15	5:8>						0000 0000	59, 67	
PCL	PC Low Byte	C Low Byte (PC<7:0>)									
TBLPTRU	_	—	bit 21	Program Mer	nory Table Poi	nter Upper By	te (TBLPTR<	20:16>)	00 0000	59, 92	
TBLPTRH	Program Mer	nory Table Poi	inter High Byt	e (TBLPTR<1	5:8>)				0000 0000	59, 92	
TBLPTRL	Program Mer	nory Table Poi	inter Low Byte	e (TBLPTR<7:	0>)				0000 0000	59, 92	
TABLAT	Program Mer	nory Table Lat	ch						0000 0000	59, 92	
PRODH	Product Regi	ster High Byte							XXXX XXXX	59, 99	
PRODL	Product Regi	ster Low Byte							XXXX XXXX	59, 99	
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INTOIE	RBIE	TMR0IF	INT0IF	RBIF	0000 000x	59, 103	
INTCON2	RBPU	INTEDG0	INTEDG1	INTEDG2	INTEDG3	TMR0IP	INT3IP	RBIP	1111 1111	59, 104	
INTCON3	INT2IP	INT1IP	INT3IE	INT2IE	INT1IE	INT3IF	INT2IF	INT1IF	1100 0000	59, 105	
INDF0	Uses content	s of FSR0 to a	address data r	nemory – valu	e of FSR0 not	changed (not	a physical re	gister)	N/A	59, 83	
POSTINC0	Uses content	s of FSR0 to a	address data r	nemory – valu	e of FSR0 pos	st-incremented	d (not a physic	al register)	N/A	59, 84	
POSTDEC0	Uses content	s of FSR0 to a	ddress data r	nemory – valu	e of FSR0 pos	st-decremente	d (not a physi	cal register)	N/A	59, 84	
PREINC0	Uses content	s of FSR0 to a	address data r	memory – valu	e of FSR0 pre	-incremented	(not a physica	al register)	N/A	59, 84	
PLUSW0	Uses content value of FSR		address data r	memory – valu	e of FSR0 pre	-incremented	(not a physica	al register) –	N/A	59, 84	
FSR0H	—	_	_	_	Indirect Data	Memory Addr	ess Pointer 0	High Byte	xxxx	59, 83	
FSR0L	Indirect Data	Memory Addre	ess Pointer 0	Low Byte					xxxx xxxx	59, 83	
WREG	Working Regi	ister							xxxx xxxx	59	
INDF1	Uses content	s of FSR1 to a	address data r	memory – valu	e of FSR1 not	changed (not	a physical re	gister)	N/A	59, 83	
POSTINC1	Uses content	s of FSR1 to a	address data r	memory – valu	e of FSR1 pos	st-incremented	d (not a physic	cal register)	N/A	59, 84	
POSTDEC1	Uses content	s of FSR1 to a	address data r	memory – valu	e of FSR1 pos	st-decremente	d (not a physi	cal register)	N/A	59, 84	
PREINC1	Uses content	s of FSR1 to a	address data r	memory – valu	e of FSR1 pre	-incremented	(not a physica	al register)	N/A	59, 84	
PLUSW1	Uses content value of FSR		address data r	memory – valu	e of FSR1 pre	-incremented	(not a physica	al register) –	N/A	59, 84	
FSR1H	—	—	—	—	Indirect Data	Memory Addr	ess Pointer 1	High Byte	xxxx	60, 83	
FSR1L	Indirect Data	Memory Addr	ess Pointer 1	Low Byte					XXXX XXXX	60, 83	
BSR	—	—	—	—	Bank Select F	Register			0000	60, 72	
INDF2	Uses content	s of FSR2 to a	address data r	memory – valu	e of FSR2 not	changed (not	a physical re	gister)	N/A	60, 83	
POSTINC2	Uses content	s of FSR2 to a	address data r	memory – valu	e of FSR2 pos	st-incremented	d (not a physic	cal register)	N/A	60, 84	
POSTDEC2	Uses content	s of FSR2 to a	address data r	memory – valu	e of FSR2 pos	st-decremente	d (not a physi	cal register)	N/A	60, 84	
PREINC2	Uses content	s of FSR2 to a	address data r	memory – valu	e of FSR2 pre	-incremented	(not a physica	al register)	N/A	60, 84	
PLUSW2		Jses contents of FSR2 to address data memory – value of FSR2 pre-incremented (not a physical register) Jses contents of FSR2 to address data memory – value of FSR2 pre-incremented (not a physical register) – ralue of FSR2 offset by W									
FSR2H		_	_	—	Indirect Data	Memory Addr	ess Pointer 2	High Byte	xxxx	60, 83	
FSR2L	Indirect Data	Memory Addr	ess Pointer 2	Low Byte					xxxx xxxx	60, 83	
STATUS	_	_	_	N	OV	Z	DC	С	x xxxx	60, 81	

 $\label{eq:Legend: Legend: Legend: u = unchanged, - = unimplemented, q = value depends on condition, r = reserved, do not modify a set of the transformation of transformation of the transformation of t$

Note 1: Bit 21 of the PC is only available in Test mode and Serial Programming modes.

2: These registers and/or bits are available only on 80-pin devices; otherwise, they are unimplemented and read as '0'. Reset states shown are for 80-pin devices.

3: Alternate names and definitions for these bits when the MSSP module is operating in I²C[™] Slave mode. See Section 18.4.3.2 "Address Masking" for details.

4: The PLLEN bit is only available in specific oscillator configurations; otherwise, it is disabled and reads as '0'. See Section 3.4.3 "PLL Frequency Multiplier" for details.

TABLE 6-3	3: PIC18F87J90 FAMILY REGISTER FILE SUMMARY (CONTINUED)									-	
File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Details on page	
TMR0H	Timer0 Regis	ter High Byte							0000 0000	60, 141	
TMR0L	Timer0 Regis	ter Low Byte							xxxx xxxx	60, 141	
T0CON	TMR00N	T08BIT	TOCS	TOSE	PSA	T0PS2	T0PS1	T0PS0	1111 1111	60, 141	
OSCCON	IDLEN	IRCF2	IRCF1	IRCF0	OSTS	IOFS	SCS1	SCS0	0110 q000	36, 60	
LCDREG	_	CPEN	BIAS2	BIAS1	BIAS0	MODE13	CKSEL1	CKSEL0	-011 1100	60, 189	
WDTCON	REGSLP	_	—	_	_	_	—	SWDTEN	00	60, 332	
RCON	IPEN	_	CM	RI	TO	PD	POR	BOR	0-11 11q0	54, 60	
TMR1H	Timer1 Regis	imer1 Register High Byte									
TMR1L	Timer1 Regis	ter Low Byte							xxxx xxxx	60, 147	
T1CON	RD16	T1RUN	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N	0000 0000	60, 143	
TMR2	Timer2 Regis	ter							0000 0000	60, 150	
PR2	Timer2 Perio	d Register							1111 1111	60, 150	
T2CON	_	T2OUTPS3	T2OUTPS2	T2OUTPS1	T2OUTPS0	TMR2ON	T2CKPS1	T2CKPS0	-000 0000	60, 149	
SSPBUF	MSSP Receiv	ve Buffer/Tran	smit Register	1	1			1	XXXX XXXX	60, 219, 254	
SSPADD	MSSP Addre	ss Register in	I ² C [™] Slave n	node. MSSP1	Baud Rate Re	eload Register	in I ² C Master	mode.	0000 0000	60, 254	
SSPSTAT	SMP	CKE	D/A	Р	S	R/W	UA	BF	0000 0000	60, 212, 221	
SSPCON1	WCOL	SSPOV	SSPEN	СКР	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000	60, 213, 222	
SSPCON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	0000 0000	60, 223,	
	GCEN	ACKSTAT	ADMSK5(3)	ADMSK4(3)	ADMSK3(3)	ADMSK2(3)	ADMSK1(3)	SEN		224	
ADRESH	A/D Result R	egister High B	yte	•	•	•	•	•	xxxx xxxx	61, 297	
ADRESL	A/D Result R	egister Low By	/te						xxxx xxxx	61, 297	
ADCON0	ADCAL	_	CHS3	CHS2	CHS1	CHS0	GO/DONE	ADON	0-00 0000	61, 289	
ADCON1	TRIGSEL	_	VCFG1	VCFG0	PCFG3	PCFG2	PCFG1	PCFG0	0-00 0000	61, 290	
ADCON2	ADFM	_	ACQT2	ACQT1	ACQT0	ADCS2	ADCS1	ADCS0	0-00 0000	61, 291	
LCDDATA4	S39C0 ⁽²⁾	S38C0 ⁽²⁾	S37C0 ⁽²⁾	S36C0 ⁽²⁾	S35C0 ⁽²⁾	S34C0 ⁽²⁾	S33C0 ⁽²⁾	S32C0	xxxx xxxx	61, 187	
LCDDATA3	S31C0	S30C0	S29C0	S28C0	S27C0	S26C0	S25C0	S24C0	xxxx xxxx	61, 187	
LCDDATA2	S23C0	S22C0	S21C0	S20C0	S19C0	S18C0	S17C0	S16C0	xxxx xxxx	61, 187	
LCDDATA1	S15C0	S14C0	S13C0	S12C0	S11C0	S10C0	S09C0	S08C0	xxxx xxxx	61, 187	
LCDDATA0	S07C0	S06C0	S05C0	S04C0	S03C0	S02C0	S01C0	S00C0	xxxx xxxx	61, 187	
LCDSE5(2)	SE47	SE46	SE45	SE44	SE43	SE42	SE41	SE40	0000 0000	61, 187	
LCDSE4	SE39 ⁽²⁾	SE38 ⁽²⁾	S37 ⁽²⁾	SE36 ⁽²⁾	SE35 ⁽²⁾	SE34 ⁽²⁾	SE33 ⁽²⁾	SE32	0000 0000	61, 187	
LCDSE3	SE31	SE30	SE29	SE28	SE27	SE26	SE25	SE24	0000 0000	61, 187	
LCDSE2	SE23	SE22	SE21	SE20	SE19	SE18	SE17	SE16	0000 0000	61, 187	
LCDSE1	SE15	SE14	SE13	SE12	SE11	SE10	SE09	SE08	0000 0000	61, 187	
CVRCON	CVREN	CVROE	CVRR	CVRSS	CVR3	CVR2	CVR1	CVR0	0000 0000	61, 305	
CMCON	C2OUT	C10UT	C2INV	C1INV	CIS	CM2	CM1	CM0	0000 0111	61, 299	
TMR3H	Timer3 Regis	ter High Byte							xxxx xxxx	61, 153	
TMR3L	Timer3 Regis	ter Low Byte							xxxx xxxx	61, 153	
T3CON	RD16	T3CCP2	T3CKPS1	T3CKPS0	T3CCP1	T3SYNC	TMR3CS	TMR30N	0000 0000	61, 151	

TABLE 6-3: PIC18F87J90 FAMILY REGISTER FILE SUMMARY (CONTINUED)

Legend: x = unknown, u = unchanged, - = unimplemented, q = value depends on condition, r = reserved, do not modify

Note 1: Bit 21 of the PC is only available in Test mode and Serial Programming modes.

2: These registers and/or bits are available only on 80-pin devices; otherwise, they are unimplemented and read as '0'. Reset states shown are for 80-pin devices.

3: Alternate names and definitions for these bits when the MSSP module is operating in I²C™ Slave mode. See Section 18.4.3.2 "Address Masking" for details.

The PLLEN bit is only available in specific oscillator configurations; otherwise, it is disabled and reads as '0'. See Section 3.4.3 "PLL Frequency Multiplier" for details.

TABLE 6-3	: PIC18F87J90 FAMILY REGISTER FILE SUMMARY (CONTINUED)									
File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Details on page
SPBRG1	EUSART Bau	ud Rate Gener	ator Low Byte	•					0000 0000	61, 259
RCREG1	EUSART Red	ceive Register							0000 0000	61, 267
TXREG1	EUSART Tra	nsmit Register	•						0000 0000	61, 265
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	0000 0010	61, 256
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000x	61, 257
LCDPS	WFT	BIASMD	LCDA	WA	LP3	LP2	LP1	LP0	0000 0000	61, 185
LCDSE0	SE07	SE06	SE05	SE04	SE03	SE02	SE01	SE00	0000 0000	61, 186
LCDCON	LCDEN	SLPEN	WERR	—	CS1	CS0	LMUX1	LMUX0	000- 0000	61, 184
EECON2	EEPROM Co	ntrol Register	2 (not a physi	cal register)						61, 90
EECON1	—	—	WPROG	FREE	WRERR	WREN	WR		00 x00-	61, 90
IPR3	—	LCDIP	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP	-111 1111	62, 114
PIR3	—	LCDIF	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF	-000 0000	62, 108
PIE3	—	LCDIE	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE	-000 0000	62, 111
IPR2	OSCFIP	CMIP	—	—	BCLIP	LVDIP	TMR3IP		11 111-	62, 113
PIR2	OSCFIF	CMIF	—	_	BCLIF	LVDIF	TMR3IF		00 000-	62, 107
PIE2	OSCFIE	CMIE	—	_	BCLIE	LVDIE	TMR3IE		00 000-	62, 110
IPR1	—	ADIP	RC1IP	TX1IP	SSPIP	_	TMR2IP	TMR1IP	-111 1-11	62, 112
PIR1	_	ADIF	RC1IF	TX1IF	SSPIF	_	TMR2IF	TMR1IF	-000 0-00	62, 106
PIE1	_	ADIE	RC1IE	TX1IE	SSPIE	_	TMR2IE	TMR1IE	-000 0-00	62, 109
OSCTUNE	INTSRC	PLLEN ⁽⁴⁾	TUN5	TUN4	TUN3	TUN2	TUN1	TUN0	0000 0000	37, 62
TRISJ ⁽²⁾	TRISJ7	TRISJ6	TRISJ5	TRISJ4	TRISJ3	TRISJ2	TRISJ1	TRISJ0	1111 1111	62, 138
TRISH ⁽²⁾	TRISH7	TRISH6	TRISH5	TRISH4	TRISH3	TRISH2	TRISH1	TRISH0	1111 1111	62, 136
TRISG	SPIOD	CCP2OD	CCP10D	TRISG4	TRISG3	TRISG2	TRISG1	TRISG0	0001 1111	62, 134
TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1		1111 111-	62, 132
TRISE	TRISE7	TRISE6	TRISE5	TRISE4	TRISE3	_	TRISE1	TRISE0	1111 1-11	62, 129
TRISD	TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0	1111 1111	62, 127
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	62, 125
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	1111 1111	62, 122
TRISA	TRISA7 ⁽⁵⁾	TRISA6 ⁽⁵⁾	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	1111 1111	62, 119
LATJ ⁽²⁾	LATJ7	LATJ6	LATJ5	LATJ4	LATJ3	LATJ2	LATJ1	LATJ0	xxxx xxxx	62, 138
LATH ⁽²⁾	LATH7	LATH6	LATH5	LATH4	LATH3	LATH2	LATH1	LATH0	xxxx xxxx	62, 136
LATG	U2OD	U10D	—	LATG4	LATG3	LATG2	LATG1	LATG0	00-x xxxx	62, 134
LATF	LATF7	LATF6	LATF5	LATF4	LATF3	LATF2	LATF1	—	xxxx xxx-	62, 132
LATE	LATE7	LATE6	LATE5	LATE4	LATE3	_	LATE1	LATE0	xxxx x-xx	62, 129
LATD	LATD7	LATD6	LATD5	LATD4	LATD3	LATD2	LATD1	LATD0	xxxx xxxx	62, 127
LATC	LATC7	LATC6	LATC5	LATC4	LATC3	LATC2	LATC1	LATC0	xxxx xxxx	62, 125
LATB	LATB7	LATB6	LATB5	LATB4	LATB3	LATB2	LATB1	LATB0	xxxx xxxx	62, 122
LATA	LATA7 ⁽⁵⁾	LATA6 ⁽⁵⁾	LATA5	LATA4	LATA3	LATA2	LATA1	LATA0	xxxx xxxx	62, 119

TABLE 6-3: PIC18F87J90 FAMILY REGISTER FILE SUMMARY (CONTINUED)

 $\label{eq:logistical_logistical$

Note 1: Bit 21 of the PC is only available in Test mode and Serial Programming modes.

2: These registers and/or bits are available only on 80-pin devices; otherwise, they are unimplemented and read as '0'. Reset states shown are for 80-pin devices.

3: Alternate names and definitions for these bits when the MSSP module is operating in I²C[™] Slave mode. See Section 18.4.3.2 "Address Masking" for details.

4: The PLLEN bit is only available in specific oscillator configurations; otherwise, it is disabled and reads as '0'. See Section 3.4.3 "PLL Frequency Multiplier" for details.

TABLE 6-3	PIC1	8F87J90	FAMILY F	REGISTE	R FILE SI	JMMARY	(CONTII	NUED)		
File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Details on page
PORTJ ⁽²⁾	RJ7	RJ6	RJ5	RJ4	RJ3	RJ2	RJ1	RJ0	xxxx xxxx	62, 138
PORTH ⁽²⁾	RH7	RH6	RH5	RH4	RH3	RH2	RH1	RH0	XXXX XXXX	62, 136
PORTG	RDPU	REPU	RJPU ⁽²⁾	RG4	RG3	RG2	RG1	RG0	000x xxxx	62, 134
PORTF	RF7	RF6	RF5	RF4	RF3	RF2	RF1	—	xxxx xxx-	62, 132
PORTE	RE7	RE6	RE5	RE4	RE3	_	RE1	RE0	xxxx x-xx	63, 129
PORTD	RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0	XXXX XXXX	63, 127
PORTC	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0	XXXX XXXX	63, 125
PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	XXXX XXXX	63, 122
PORTA	RA7 ⁽⁵⁾	RA6 ⁽⁵⁾	RA5	RA4	RA3	RA2	RA1	RA0	xx0x 0000	63, 119
SPBRGH1	EUSART Bau	d Rate Gener	ator High Byte	e					0000 0000	63, 259
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN	0100 0-00	63, 258
LCDDATA23 ⁽²⁾	S47C3	S46C3	S45C3	S44C3	S43C3	S42C3	S41C3	S40C3	xxxx xxxx	63, 187
LCDDATA22	S39C3 ⁽²⁾	S38C3 ⁽²⁾	S37C3 ⁽²⁾	S36C3 ⁽²⁾	S35C3 ⁽²⁾	S34C3 ⁽²⁾	S33C3 ⁽²⁾	S32C3	xxxx xxxx	63, 187
LCDDATA21	S31C3	S30C3	S29C3	S28C3	S27C3	S26C3	S25C3	S24C3	xxxx xxxx	63, 187
LCDDATA20	S23C3	S22C3	S21C3	S20C3	S19C3	S18C3	S17C3	S16C3	xxxx xxxx	63, 187
LCDDATA19	S15C3	S14C3	S13C3	S12C3	S11C3	S10C3	S09C3	S08C3	xxxx xxxx	63, 187
LCDDATA18	S07C3	S06C3	S05C3	S04C3	S03C3	S02C3	S01C3	S00C3	xxxx xxxx	63, 187
LCDDATA17 ⁽²⁾	S47C2	S46C2	S45C2	S44C2	S43C2	S42C2	S41C2	S40C2	xxxx xxxx	63, 187
LCDDATA16	S39C2 ⁽²⁾	S38C2 ⁽²⁾	S37C2 ⁽²⁾	S36C2 ⁽²⁾	S35C2 ⁽²⁾	S34C2 ⁽²⁾	S33C2 ⁽²⁾	S32C2	xxxx xxxx	63, 187
LCDDATA15	S31C2	S30C2	S29C2	S28C2	S27C2	S26C2	S25C2	S24C2	xxxx xxxx	63, 187
LCDDATA14	S23C2	S22C2	S21C2	S20C2	S19C2	S18C2	S17C2	S16C2	xxxx xxxx	63, 187
LCDDATA13	S15C2	S14C2	S13C2	S12C2	S11C2	S10C2	S09C2	S08C2	xxxx xxxx	63, 187
LCDDATA12	S07C2	S06C2	S05C2	S04C2	S03C2	S02C2	S01C2	S00C2	xxxx xxxx	63, 187
LCDDATA11 ⁽²⁾	S47C1	S46C1	S45C1	S44C1	S43C1	S42C1	S41C1	S40C1	xxxx xxxx	63, 187
LCDDATA10	S39C1 ⁽²⁾	S38C1 ⁽²⁾	S37C1 ⁽²⁾	S36C1 ⁽²⁾	S35C1 ⁽²⁾	S34C1 ⁽²⁾	S33C1 ⁽²⁾	S32C1	xxxx xxxx	63, 187
LCDDATA9	S31C1	S30C1	S29C1	S28C1	S27C1	S26C1	S25C1	S24C1	xxxx xxxx	63, 187
LCDDATA8	S23C1	S22C1	S21C1	S20C1	S19C1	S18C1	S17C1	S16C1	xxxx xxxx	63, 187
LCDDATA7	S15C1	S14C1	S13C1	S12C1	S11C1	S10C1	S09C1	S08C1	xxxx xxxx	63, 187
LCDDATA6	S07C1	S06C1	S05C1	S04C1	S03C1	S02C1	S01C1	S00C1	xxxx xxxx	63, 187
LCDDATA5 ⁽²⁾	S47C0	S46C0	S45C0	S44C0	S43C0	S42C0	S41C0	S40C0	xxxx xxxx	63, 187
CCPR1H	Capture/Com	pare/PWM Re	gister 1 High	Byte		11			xxxx xxxx	63, 174
CCPR1L	· ·	pare/PWM Re							xxxx xxxx	63, 174
CCP1CON	· _	_	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	00 0000	63, 173
CCPR2H	Capture/Com	pare/PWM Re	aister 2 High	Byte	1				xxxx xxxx	63, 174
CCPR2L		pare/PWM Re							xxxx xxxx	64, 174
CCP2CON	_	_	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	00 0000	64, 173
SPBRG2	AUSART Bau	I Id Rate Gener							0000 0000	64, 278
RCREG2		ceive Register							0000 0000	64, 283
TXREG2		nsmit Register							0000 0000	64, 281
TXSTA2	CSRC	TX9	TXEN	SYNC		BRGH	TRMT	TX9D	0000 -010	64, 276
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000x	
RTCCFG	RTCEN		RTCWREN	RTCSYNC	HALFSEC	RTCOE	RTCPTR1	RTCPTR0	0-00 0000	64, 157
RTCCAL	CAL7	CAL6	CAL5	CAL4	CAL3	CAL2	CAL1	CAL0	0000 0000	64, 158
RTCVALH		High Register				.	с, ш I	0. 120	xxxx xxxx	64, 160
		<u> </u>								01,100

TABLE 6-3: PIC18F87J90 FAMILY REGISTER FILE SUMMARY (CONTINUED)

Legend: x = unknown, u = unchanged, - = unimplemented, q = value depends on condition, r = reserved, do not modify

Note 1: Bit 21 of the PC is only available in Test mode and Serial Programming modes.

2: These registers and/or bits are available only on 80-pin devices; otherwise, they are unimplemented and read as '0'. Reset states shown are for 80-pin devices.

3: Alternate names and definitions for these bits when the MSSP module is operating in I²C[™] Slave mode. See Section 18.4.3.2 "Address Masking" for details.

4: The PLLEN bit is only available in specific oscillator configurations; otherwise, it is disabled and reads as '0'. See Section 3.4.3 "PLL Frequency Multiplier" for details.

TABLE 6-3: PIC18F87J90 FAMILY REGISTER FILE SUMMARY (CONTINUED)

								/		
File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Details on page
RTCVALL	RTCC Value	XXXX XXXX	64, 160							
ALRMCFG	ALRMEN	ALRMPTR0	0000 0000	64, 159						
ALRMRPT	ARPT7	ARPT6	ARPT5	ARPT4	ARPT3	ARPT2	ARPT1	ARPT0	0000 0000	64, 160
ALRMVALH	Alarm Value H	High Register	Window based	d on ALRMPT	R<1:0>				XXXX XXXX	64, 163
ALRMVALL	Alarm Value L	Low Register \	Vindow based	on ALRMPT	R<1:0>				XXXX XXXX	64, 163
CTMUCONH	CTMUEN	_	CTMUSIDL	TGEN	EDGEN	EDGSEQEN	IDISSEN	CTTRIG	0-00 0000	64, 321
CTMUCONL	EDG2POL	EDG2SEL1	EDG2SEL0	EDG1POL	EDG1SEL1	EDG1SEL0	EDG2STAT	EDG1STAT	0000 0000	64, 322
CTMUICON	ITRIM5	ITRIM4	ITRIM3	ITRIM2	ITRIM1	ITRIM0	IRNG1	IRNG0	0000 0000	64, 323
PADCFG1	—	_	_	_	_	RTSECSEL1	RTSECSEL0	_	00-	64, 158

Legend: x = unknown, u = unchanged, - = unimplemented, q = value depends on condition, r = reserved, do not modify

Note 1: Bit 21 of the PC is only available in Test mode and Serial Programming modes.

2: These registers and/or bits are available only on 80-pin devices; otherwise, they are unimplemented and read as '0'. Reset states shown are for 80-pin devices.

3: Alternate names and definitions for these bits when the MSSP module is operating in I^2C^{TM} Slave mode. See Section 18.4.3.2 "Address Masking" for details.

4: The PLLEN bit is only available in specific oscillator configurations; otherwise, it is disabled and reads as '0'. See Section 3.4.3 "PLL Frequency Multiplier" for details.

6.3.5 STATUS REGISTER

The STATUS register, shown in Register 6-2, contains the arithmetic status of the ALU. The STATUS register can be the operand for any instruction, as with any other register. If the STATUS register is the destination for an instruction that affects the Z, DC, C, OV or N bits, then the write to these five bits is disabled.

These bits are set or cleared according to the device logic. Therefore, the result of an instruction with the STATUS register as destination may be different than intended. For example, CLRF STATUS will set the Z bit but leave the other bits unchanged. The STATUS

register then reads back as '000u u1uu'. It is recommended, therefore, that only BCF, BSF, SWAPF, MOVFF and MOVWF instructions are used to alter the STATUS register because these instructions do not affect the Z, C, DC, OV or N bits in the STATUS register.

For other instructions not affecting any Status bits, see the instruction set summaries in Table 26-2 and Table 26-3.

Note: The C and DC bits operate as a borrow and digit borrow bit, respectively, in subtraction.

REGISTER 6-2: STATUS REGISTER

U-0	U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
	—	—	N	OV	Z	DC ⁽¹⁾	C ⁽²⁾
bit 7							bit (
Legend:							
R = Read	able bit	W = Writable	bit	U = Unimplem	nented bit, rea	ad as '0'	
-n = Value	e at POR	'1' = Bit is set		'0' = Bit is clea		x = Bit is unkn	iown
bit 7-5	Unimpleme	ented: Read as '	כי				
bit 4	N: Negative	bit					
		sed for signed ar LU MSB = 1).	ithmetic (2's co	omplement). It i	ndicates whe	ther the result wa	as
	1 = Result w 0 = Result w	vas negative vas positive					
bit 3	OV: Overflor	w bit					
		sed for signed ar				verflow of the	
	•	ude which cause	•	. , .		`	
		v occurred for sig	gned anthmetio	c (in this anthree	ette operation)	
bit 2	Z: Zero bit						
		ult of an arithmet ult of an arithmet			0		
bit 1		arry/Borrow bit ⁽¹⁾					
	•	ADDLW, SUBL		instructions:			
		out from the 4th			urred		
		/-out from the 4th	n low-order bit	of the result			
bit 0	C: Carry/Bo	ADDLW, SUBL	w and SUDWE	netructione:			
		out from the Mos			courred		
		/-out from the Mo					
Note 1:		polarity is reverse ate (RRF,RLF)					
2:	For borrow, the	polarity is reverse	ed. A subtracti	on is executed b	by adding the	2's complement	of the second
	operand. For rot source register.	tate (RRF, RLF)	instructions, t	his bit is loaded	with either th	e high or low-or	der bit of the

6.4 Data Addressing Modes

Note:	The execution of some instructions in the
	core PIC18 instruction set are changed
	when the PIC18 extended instruction set is
	enabled. See Section 6.6 "Data Memory
	and the Extended Instruction Set" for
	more information.

While the program memory can be addressed in only one way, through the program counter, information in the data memory space can be addressed in several ways. For most instructions, the addressing mode is fixed. Other instructions may use up to three modes, depending on which operands are used and whether or not the extended instruction set is enabled.

The addressing modes are:

- Inherent
- Literal
- Direct
- Indirect

An additional addressing mode, Indexed Literal Offset, is available when the extended instruction set is enabled (XINST Configuration bit = 1). Its operation is discussed in greater detail in **Section 6.6.1 "Indexed Addressing with Literal Offset**".

6.4.1 INHERENT AND LITERAL ADDRESSING

Many PIC18 control instructions do not need any argument at all; they either perform an operation that globally affects the device or they operate implicitly on one register. This addressing mode is known as Inherent Addressing. Examples include: SLEEP, RESET and DAW.

Other instructions work in a similar way, but require an additional explicit argument in the opcode. This is known as Literal Addressing mode, because they require some literal value as an argument. Examples include: ADDLW and MOVLW, which respectively, add or move a literal value to the W register. Other examples include CALL and GOTO, which include a 20-bit program memory address.

6.4.2 DIRECT ADDRESSING

Direct Addressing specifies all or part of the source and/or destination address of the operation within the opcode itself. The options are specified by the arguments accompanying the instruction.

In the core PIC18 instruction set, bit-oriented and byte-oriented instructions use some version of Direct Addressing by default. All of these instructions include some 8-bit literal address as their Least Significant Byte. This address specifies either a register address in one of the banks of data RAM (Section 6.3.3 "General Purpose Register File") or a location in the Access Bank (Section 6.3.2 "Access Bank") as the data source for the instruction. The Access RAM bit, 'a', determines how the address is interpreted. When 'a' is '1', the contents of the BSR (Section 6.3.1 "Bank Select Register") are used with the address to determine the complete 12-bit address of the register. When 'a' is '0', the address is interpreted as being a register in the Access Bank. Addressing that uses the Access RAM is sometimes also known as Direct Forced Addressing mode.

A few instructions, such as MOVFF, include the entire 12-bit address (either source or destination) in their opcodes. In these cases, the BSR is ignored entirely.

The destination of the operation's results is determined by the destination bit, 'd'. When 'd' is '1', the results are stored back in the source register, overwriting its original contents. When 'd' is '0', the results are stored in the W register. Instructions without the 'd' argument have a destination that is implicit in the instruction; their destination is either the target register being operated on or the W register.

6.4.3 INDIRECT ADDRESSING

Indirect Addressing allows the user to access a location in data memory without giving a fixed address in the instruction. This is done by using File Select Registers (FSRs) as pointers to the locations to be read or written to. Since the FSRs are themselves located in RAM as Special Function Registers, they can also be directly manipulated under program control. This makes FSRs very useful in implementing data structures, such as tables and arrays in data memory.

The registers for Indirect Addressing are also implemented with Indirect File Operands (INDFs) that permit automatic manipulation of the pointer value with auto-incrementing, auto-decrementing or offsetting with another value. This allows for efficient code using loops, such as the example of clearing an entire RAM bank in Example 6-5. It also enables users to perform Indexed Addressing and other Stack Pointer operations for program memory in data memory.

EXAMPLE 6-5: HOW TO CLEAR RAM (BANK 1) USING INDIRECT ADDRESSING

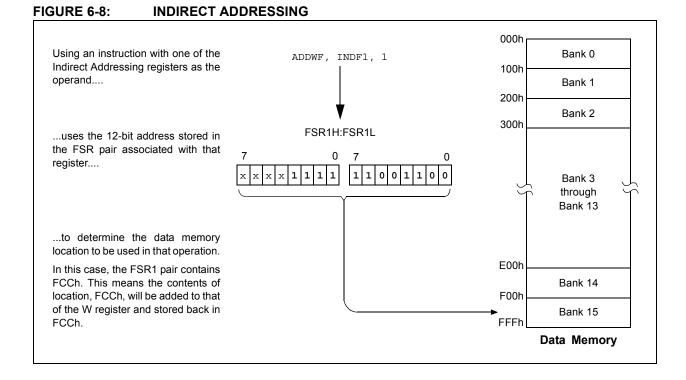
	LFSR	FSR0, 100h	;	
NEXT	CLRF	POSTINC0	;	Clear INDF
			;	register then
			;	inc pointer
	BTFSS	FSROH, 1	;	All done with
			;	Bank1?
	BRA	NEXT	;	NO, clear next
CONTIN	UE		;	YES, continue

6.4.3.1 FSR Registers and the INDF Operand

At the core of Indirect Addressing are three sets of registers: FSR0, FSR1 and FSR2. Each represents a pair of 8-bit registers, FSRnH and FSRnL. The four upper bits of the FSRnH register are not used, so each FSR pair holds a 12-bit value. This represents a value that can address the entire range of the data memory in a linear fashion. The FSR register pairs, then, serve as pointers to data memory locations.

Indirect Addressing is accomplished with a set of Indirect File Operands, INDF0 through INDF2. These can be thought of as "virtual" registers: they are mapped in the SFR space but are not physically implemented. Reading or writing to a particular INDF register actually accesses its corresponding FSR register pair. A read from INDF1, for example, reads the data at the address indicated by FSR1H:FSR1L. Instructions that use the INDF registers as operands actually use the contents of their corresponding FSR as a pointer to the instruction's target. The INDF operand is just a convenient way of using the pointer.

Because Indirect Addressing uses a full, 12-bit address, data RAM banking is not necessary. Thus, the current contents of the BSR and the Access RAM bit have no effect on determining the target address.



6.4.3.2 FSR Registers and POSTINC, POSTDEC, PREINC and PLUSW

In addition to the INDF operand, each FSR register pair also has four additional indirect operands. Like INDF, these are "virtual" registers that cannot be indirectly read or written to. Accessing these registers actually accesses the associated FSR register pair, but also performs a specific action on its stored value. They are:

- POSTDEC: accesses the FSR value, then automatically decrements it by '1' afterwards
- POSTINC: accesses the FSR value, then automatically increments it by '1' afterwards
- PREINC: increments the FSR value by '1', then uses it in the operation
- PLUSW: adds the signed value of the W register (range of -127 to 128) to that of the FSR and uses the new value in the operation

In this context, accessing an INDF register uses the value in the FSR registers without changing them. Similarly, accessing a PLUSW register gives the FSR value offset by the value in the W register; neither value is actually changed in the operation. Accessing the other virtual registers changes the value of the FSR registers.

Operations on the FSRs with POSTDEC, POSTINC and PREINC affect the entire register pair; that is, rollovers of the FSRnL register from FFh to 00h carry over to the FSRnH register. On the other hand, results of these operations do not change the value of any flags in the STATUS register (e.g., Z, N, OV, etc.).

The PLUSW register can be used to implement a form of Indexed Addressing in the data memory space. By manipulating the value in the W register, users can reach addresses that are fixed offsets from pointer addresses. In some applications, this can be used to implement some powerful program control structure, such as software stacks, inside of data memory.

6.4.3.3 Operations by FSRs on FSRs

Indirect Addressing operations that target other FSRs or virtual registers represent special cases. For example, using an FSR to point to one of the virtual registers will not result in successful operations. As a specific case, assume that the FSR0H:FSR0L registers contain FE7h, the address of INDF1. Attempts to read the value of INDF1, using INDF0 as an operand, will return 00h. Attempts to write to INDF1, using INDF0 as the operand, will result in a NOP.

On the other hand, using the virtual registers to write to an FSR pair may not occur as planned. In these cases, the value will be written to the FSR pair but without any incrementing or decrementing. Thus, writing to INDF2 or POSTDEC2 will write the same value to the FSR2H:FSR2L registers.

Since the FSRs are physical registers mapped in the SFR space, they can be manipulated through all direct operations. Users should proceed cautiously when working on these registers, particularly if their code uses Indirect Addressing.

Similarly, operations by Indirect Addressing are generally permitted on all other SFRs. Users should exercise the appropriate caution that they do not inadvertently change settings that might affect the operation of the device.

6.5 Program Memory and the Extended Instruction Set

The operation of program memory is unaffected by the use of the extended instruction set.

Enabling the extended instruction set adds five additional two-word commands to the existing PIC18 instruction set: ADDFSR, CALLW, MOVSF, MOVSS and SUBFSR. These instructions are executed as described in Section 6.2.4 "Two-Word Instructions".

6.6 Data Memory and the Extended Instruction Set

Enabling the PIC18 extended instruction set (XINST Configuration bit = 1) significantly changes certain aspects of data memory and its addressing. Specifically, the use of the Access Bank for many of the core PIC18 instructions is different; this is due to the introduction of a new addressing mode for the data memory space. This mode also alters the behavior of Indirect Addressing using FSR2 and its associated operands.

What does not change is just as important. The size of the data memory space is unchanged, as well as its linear addressing. The SFR map remains the same. Core PIC18 instructions can still operate in both Direct and Indirect Addressing modes. Inherent and literal instructions do not change at all. Indirect Addressing with FSR0 and FSR1 also remains unchanged.

6.6.1 INDEXED ADDRESSING WITH LITERAL OFFSET

Enabling the PIC18 extended instruction set changes the behavior of Indirect Addressing using the FSR2 register pair and its associated file operands. Under the proper conditions, instructions that use the Access Bank – that is, most bit-oriented and byte-oriented instructions – can invoke a form of Indexed Addressing using an offset specified in the instruction. This special addressing mode is known as Indexed Addressing with Literal Offset, or Indexed Literal Offset mode.

When using the extended instruction set, this addressing mode requires the following:

- The use of the Access Bank is forced ('a' = 0); and
- The file address argument is less than or equal to 5Fh.

Under these conditions, the file address of the instruction is not interpreted as the lower byte of an address (used with the BSR in Direct Addressing) or as an 8-bit address in the Access Bank. Instead, the value is interpreted as an offset value to an Address Pointer specified by FSR2. The offset and the contents of FSR2 are added to obtain the target address of the operation.

6.6.2 INSTRUCTIONS AFFECTED BY INDEXED LITERAL OFFSET MODE

Any of the core PIC18 instructions that can use Direct Addressing are potentially affected by the Indexed Literal Offset Addressing mode. This includes all byte-oriented and bit-oriented instructions or almost one-half of the standard PIC18 instruction set. Instructions that only use Inherent or Literal Addressing modes are unaffected.

Additionally, byte-oriented and bit-oriented instructions are not affected if they do not use the Access Bank (Access RAM bit is '1') or include a file address of 60h or above. Instructions meeting these criteria will continue to execute as before. A comparison of the different possible addressing modes when the extended instruction set is enabled is shown in Figure 6-9.

Those who desire to use byte-oriented or bit-oriented instructions in the Indexed Literal Offset mode should note the changes to assembler syntax for this mode. This is described in more detail in **Section 26.2.1** "Extended Instruction Syntax".

FIGURE 6-9: COMPARING ADDRESSING OPTIONS FOR BIT-ORIENTED AND BYTE-ORIENTED INSTRUCTIONS (EXTENDED INSTRUCTION SET ENABLED)

EXAMPLE INSTRUCTION: ADDWF, f, d, a (Opcode: 0010 01da ffff ffff)

When a = 0 and $f \ge 60h$:

The instruction executes in Direct Forced mode. 'f' is interpreted as a location in the Access RAM between 060h and FFFh. This is the same as locations, F60h to FFFh (Bank 15), of data memory.

Locations below 060h are not available in this addressing mode.

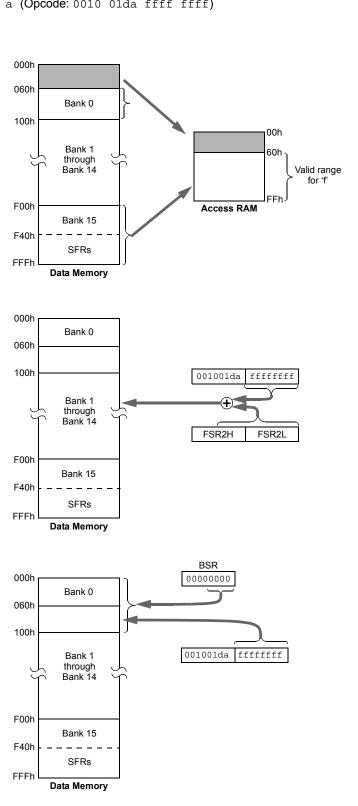
When a = 0 and $f \le 5Fh$:

The instruction executes in Indexed Literal Offset mode. 'f' is interpreted as an offset to the address value in FSR2. The two are added together to obtain the address of the target register for the instruction. The address can be anywhere in the data memory space.

Note that in this mode, the correct syntax is now: ADDWF [k], d where 'k' is the same as 'f'.

When a = 1 (all values of f):

The instruction executes in Direct mode (also known as Direct Long mode). 'f' is interpreted as a location in one of the 16 banks of the data memory space. The bank is designated by the Bank Select Register (BSR). The address can be in any implemented bank in the data memory space.



6.6.3 MAPPING THE ACCESS BANK IN INDEXED LITERAL OFFSET MODE

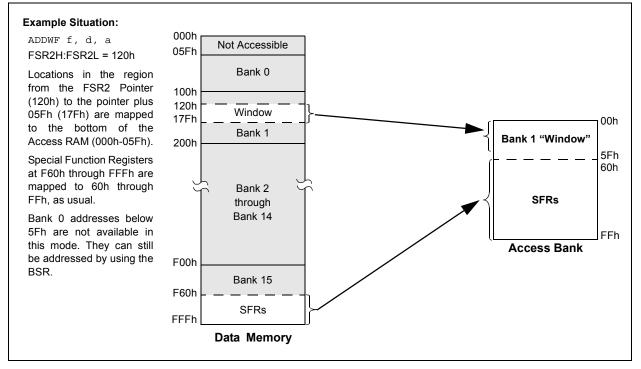
The use of Indexed Literal Offset Addressing mode effectively changes how the lower part of Access RAM (00h to 5Fh) is mapped. Rather than containing just the contents of the bottom part of Bank 0, this mode maps the contents from Bank 0 and a user-defined "window" that can be located anywhere in the data memory space. The value of FSR2 establishes the lower boundary of the addresses mapped into the window, while the upper boundary is defined by FSR2 plus 95 (5Fh). Addresses in the Access RAM above 5Fh are mapped as previously described (see Section 6.3.2 "Access Bank"). An example of Access Bank remapping in this addressing mode is shown in Figure 6-10.

Remapping of the Access Bank applies *only* to operations using the Indexed Literal Offset mode. Operations that use the BSR (Access RAM bit is '1') will continue to use Direct Addressing as before. Any Indirect or Indexed Addressing operation that explicitly uses any of the indirect file operands (including FSR2) will continue to operate as standard Indirect Addressing. Any instruction that uses the Access Bank, but includes a register address of greater than 05Fh, will use Direct Addressing and the normal Access Bank map.

6.6.4 BSR IN INDEXED LITERAL OFFSET MODE

Although the Access Bank is remapped when the extended instruction set is enabled, the operation of the BSR remains unchanged. Direct Addressing, using the BSR to select the data memory bank, operates in the same manner as previously described.

FIGURE 6-10: REMAPPING THE ACCESS BANK WITH INDEXED LITERAL OFFSET ADDRESSING



NOTES:

7.0 FLASH PROGRAM MEMORY

The Flash program memory is readable, writable and erasable during normal operation over the entire VDD range.

A read from program memory is executed on one byte at a time. A write to program memory is executed on blocks of 64 bytes at a time or two bytes at a time. Program memory is erased in blocks of 1024 bytes at a time. A bulk erase operation may not be issued from user code.

Writing or erasing program memory will cease instruction fetches until the operation is complete. The program memory cannot be accessed during the write or erase, therefore, code cannot execute. An internal programming timer terminates program memory writes and erases.

A value written to program memory does not need to be a valid instruction. Executing a program memory location that forms an invalid instruction results in a NOP.

7.1 Table Reads and Table Writes

In order to read and write program memory, there are two operations that allow the processor to move bytes between the program memory space and the data RAM:

- Table Read (TBLRD)
- Table Write (TBLWT)

The program memory space is 16 bits wide, while the data RAM space is 8 bits wide. Table reads and table writes move data between these two memory spaces through an 8-bit register (TABLAT).

Table read operations retrieve data from program memory and place it into the data RAM space. Figure 7-1 shows the operation of a table read with program memory and data RAM.

Table write operations store data from the data memory space into holding registers in program memory. The procedure to write the contents of the holding registers into program memory is detailed in **Section 7.5 "Writing to Flash Program Memory"**. Figure 7-2 shows the operation of a table write with program memory and data RAM.

Table operations work with byte entities. A table block containing data, rather than program instructions, is not required to be word-aligned. Therefore, a table block can start and end at any byte address. If a table write is being used to write executable code into program memory, program instructions will need to be word-aligned.

FIGURE 7-1: TABLE READ OPERATION

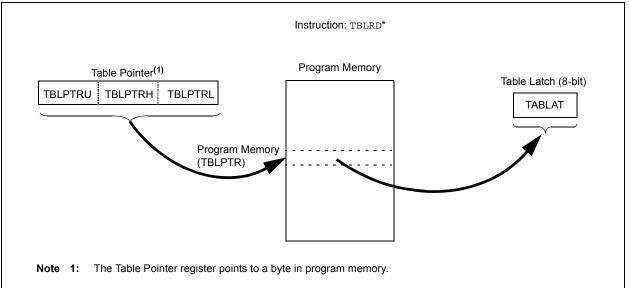
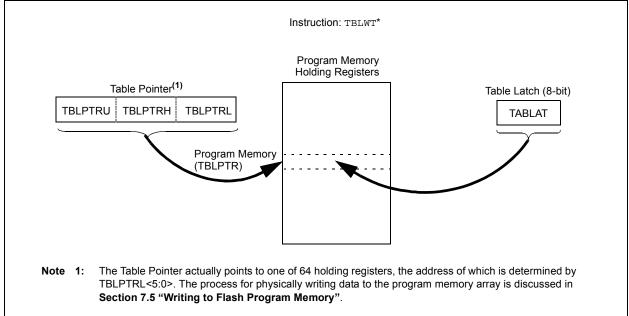


FIGURE 7-2: TABLE WRITE OPERATION



7.2 Control Registers

Several control registers are used in conjunction with the TBLRD and TBLWT instructions. These include the:

- EECON1 register
- EECON2 register
- TABLAT register
- TBLPTR registers

7.2.1 EECON1 AND EECON2 REGISTERS

The EECON1 register (Register 7-1) is the control register for memory accesses. The EECON2 register is not a physical register; it is used exclusively in the memory write and erase sequences. Reading EECON2 will read all '0's.

The WPROG bit, when set, allows the user to program a single word (two bytes) upon the execution of the WR command. If this bit is cleared, the WR command programs a block of 64 bytes. The FREE bit, when set, will allow a program memory erase operation. When FREE is set, the erase operation is initiated on the next WR command. When FREE is clear, only writes are enabled.

The WREN bit, when set, will allow a write operation. On power-up, the WREN bit is clear. The WRERR bit is set in hardware when the WR bit is set, and cleared when the internal programming timer expires and the write operation is complete.

Note:	During normal operation, the WRERR is
	read as '1'. This can indicate that a write
	operation was prematurely terminated by
	a Reset or a write operation was
	attempted improperly.

The WR control bit initiates write operations. The bit cannot be cleared, only set, in software. It is cleared in hardware at the completion of the write operation.

U-0	U-0	R/W-0	R/W-0	R/W-x	R/W-0	R/S-0	U-0
	_	WPROG	FREE	WRERR ⁽¹⁾	WREN	WR	
bit 7							bit
Legend:		S = Settable b	it (cannot be	cleared in softwa	are)		
R = Reada	ble bit	W = Writable	oit	U = Unimplem	ented bit, read	d as '0'	
-n = Value	at POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkn	own
bit 7-6	Unimplem	nented: Read as 'o)'				
bit 5	WPROG:	One Word-Wide P	rogram bit				
	•	am 2 bytes on the					
	0	am 64 bytes on the		mmand			
bit 4		ish Erase Enable b					
		rms an erase opera rm write-only	ation on the ne	ext WR command	d (cleared by c	ompletion of era	ase operation
bit 3	WRERR:	Flash Program Err	or Flag bit ⁽¹⁾				
		e operation is prer tion, or an imprope			et during self-t	timed programm	ning in norma
		rite operation com		50)			
bit 2		ash Program Write					
		s write cycles to FI		memorv			
		ts write cycles to F					
bit 1	WR: Write	Control bit					
	(The c can o	es a program mem operation is self-tin nly be set (not clea	ned and the bi	it is cleared by h	ardware once	write is complet	e. The WR b
		cycle is complete					
bit 0	Unimplem	nented: Read as 'o)'				
							c

REGISTER 7-1: EECON1: EEPROM CONTROL REGISTER 1

Note 1: When a WRERR error occurs, the EEPGD and CFGS bits are not cleared. This allows tracing of the error condition.

7.2.2 TABLE LATCH REGISTER (TABLAT)

The Table Latch (TABLAT) is an 8-bit register mapped into the SFR space. The Table Latch register is used to hold 8-bit data during data transfers between program memory and data RAM.

7.2.3 TABLE POINTER REGISTER (TBLPTR)

The Table Pointer (TBLPTR) register addresses a byte within the program memory. The TBLPTR is comprised of three SFR registers: Table Pointer Upper Byte, Table Pointer High Byte and Table Pointer Low Byte (TBLPTRU:TBLPTRH:TBLPTRL). These three registers join to form a 22-bit wide pointer. The low-order 21 bits allow the device to address up to 2 Mbytes of program memory space. The 22nd bit allows access to the Device ID, the user ID and the Configuration bits.

The Table Pointer register, TBLPTR, is used by the TBLRD and TBLWT instructions. These instructions can update the TBLPTR in one of four ways based on the table operation. These operations are shown in Table 7-1. These operations on the TBLPTR only affect the low-order 21 bits.

7.2.4 TABLE POINTER BOUNDARIES

TBLPTR is used in reads, writes and erases of the Flash program memory.

When a TBLRD is executed, all 22 bits of the TBLPTR determine which byte is read from program memory into TABLAT.

When a TBLWT is executed, the seven LSbs of the Table Pointer register (TBLPTR<6:0>) determine which of the 64 program memory holding registers is written to. When the timed write to program memory begins (via the WR bit), the 12 MSbs of the TBLPTR (TBLPTR<21:10>) determine which program memory block of 1024 bytes is written to. For more detail, see **Section 7.5 "Writing to Flash Program Memory"**.

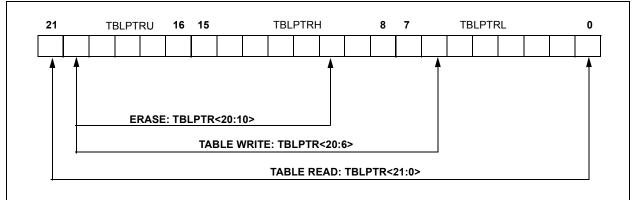
When an erase of program memory is executed, the 12 MSbs of the Table Pointer register point to the 1024-byte block that will be erased. The Least Significant bits are ignored.

Figure 7-3 describes the relevant boundaries of TBLPTR based on Flash program memory operations.

TABLE 7-1: TABLE POINTER OPERATIONS WITH TBLRD AND TBLWT INSTRUCTIONS

Example	Operation on Table Pointer
TBLRD* TBLWT*	TBLPTR is not modified
TBLRD*+ TBLWT*+	TBLPTR is incremented after the read/write
TBLRD*- TBLWT*-	TBLPTR is decremented after the read/write
TBLRD+* TBLWT+*	TBLPTR is incremented before the read/write

FIGURE 7-3: TABLE POINTER BOUNDARIES BASED ON OPERATION



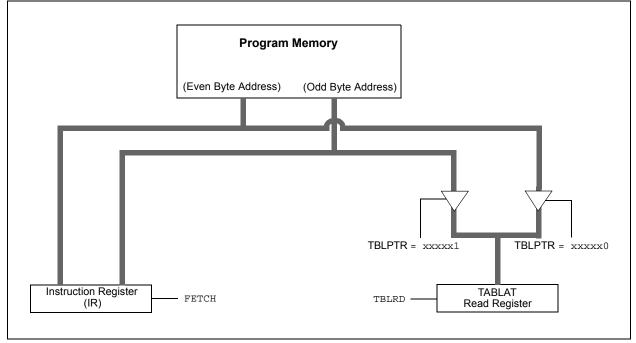
7.3 Reading the Flash Program Memory

The TBLRD instruction is used to retrieve data from program memory and places it into data RAM. Table reads from program memory are performed one byte at a time.

TBLPTR points to a byte address in program space. Executing TBLRD places the byte pointed to into TABLAT. In addition, TBLPTR can be modified automatically for the next table read operation.

The internal program memory is typically organized by words. The Least Significant bit of the address selects between the high and low bytes of the word. Figure 7-4 shows the interface between the internal program memory and the TABLAT.

FIGURE 7-4: READS FROM FLASH PROGRAM MEMORY



EXAMPLE 7-1: READING A FLASH PROGRAM MEMORY WORD

	MOVLW MOVWF	CODE_ADDR_UPPER TBLPTRU		Load TBLPTR with the base address of the word
	MOVUF	CODE ADDR HIGH	'	address of the word
	MOVWF	TBLPTRH		
	MOVLW	CODE_ADDR_LOW		
	MOVWF	TBLPTRL		
READ_WORD				
	TBLRD*+		;	read into TABLAT and increment
	MOVF	TABLAT, W	;	get data
	MOVWF	WORD_EVEN		
	TBLRD*+		;	read into TABLAT and increment
	MOVF	TABLAT, W	;	get data
	MOVWF	WORD_ODD		

7.4 Erasing Flash Program Memory

The minimum erase block is 512 words or 1024 bytes. Only through the use of an external programmer, or through ICSP control, can larger blocks of program memory be bulk erased. Word erase in the Flash array is not supported.

When initiating an erase sequence from the microcontroller itself, a block of 1024 bytes of program memory is erased. The Most Significant 12 bits of the TBLPTR<21:10> point to the block being erased. The TBLPTR<9:0> bits are ignored.

The EECON1 register commands the erase operation. The WREN bit must be set to enable write operations. The FREE bit is set to select an erase operation. For protection, the write initiate sequence for EECON2 must be used.

A long write is necessary for erasing the internal Flash. Instruction execution is halted while in a long write cycle. The long write will be terminated by the internal programming timer.

7.4.1 FLASH PROGRAM MEMORY ERASE SEQUENCE

The sequence of events for erasing a block of internal program memory location is:

- 1. Load Table Pointer register with address being erased.
- 2. Set the WREN and FREE bits (EECON1<2,4>) to enable the erase operation.
- 3. Disable interrupts.
- 4. Write 55h to EECON2.
- 5. Write 0AAh to EECON2.
- 6. Set the WR bit. This will begin the erase cycle.
- 7. The CPU will stall for the duration of the erase for TIW (see parameter D133B).
- 8. Re-enable interrupts.

EXAMPLE 7-2: ERASING FLASH PROGRAM MEMORY

	MOVLW MOVWF MOVLW MOVWF MOVLW	CODE_ADDR_UPPER TBLPTRU CODE_ADDR_HIGH TBLPTRH CODE_ADDR_LOW	; load TBLPTR with the base ; address of the memory block
	MOVWF	TBLPTRL	
ERASE			
	BSF	EECON1, WREN	
	BSF	EECON1, FREE	; enable Erase operation
	BCF	INTCON, GIE	; disable interrupts
Required	MOVLW	55h	
Sequence	MOVWF	EECON2	; write 55h
	MOVLW	0AAh	
	MOVWF	EECON2	; write OAAh
	BSF	EECON1, WR	; start erase (CPU stall)
	BSF	INTCON, GIE	; re-enable interrupts
			-

7.5 Writing to Flash Program Memory

The programming block is 32 words or 64 bytes. Programming one word or two bytes at a time is also supported.

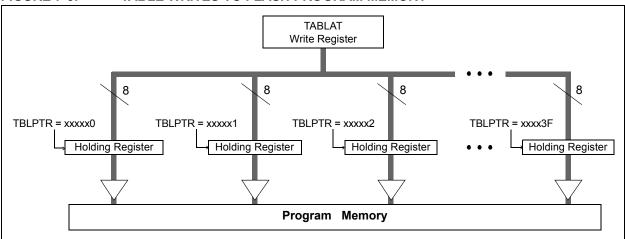
Table writes are used internally to load the holding registers needed to program the Flash memory. There are 64 holding registers used by the table writes for programming.

Since the Table Latch (TABLAT) is only a single byte, the TBLWT instruction may need to be executed 64 times for each programming operation (if WPROG = 0). All of the table write operations will essentially be short writes because only the holding registers are written. At the end of updating the 64 holding registers, the EECON1 register must be written to in order to start the programming operation with a long write.

The long write is necessary for programming the internal Flash. Instruction execution is halted while in a long write cycle. The long write will be terminated by the internal programming timer. The on-chip timer controls the write time. The write/erase voltages are generated by an on-chip charge pump, rated to operate over the voltage range of the device.

- Note 1: Unlike previous PIC18 Flash devices, members of the PIC18F87J90 family do not reset the holding registers after a write occurs. The holding registers must be cleared or overwritten before a programming sequence.
 - 2: To maintain the endurance of the program memory cells, each Flash byte should not be programmed more than one time between erase operations. Before attempting to modify the contents of the target cell a second time, an erase of the target, or a bulk erase of the entire memory, must be performed.





7.5.1 FLASH PROGRAM MEMORY WRITE SEQUENCE

The sequence of events for programming an internal program memory location should be:

- 1. Read 1024 bytes into RAM.
- 2. Update data values in RAM as necessary.
- 3. Load Table Pointer register with the address being erased.
- 4. Execute the erase procedure.
- 5. Load the Table Pointer register with the address of the first byte being written, minus 1.
- 6. Write the 64 bytes into the holding registers with auto-increment.
- 7. Set the WREN bit (EECON1<2>) to enable byte writes.

- 8. Disable interrupts.
- 9. Write 55h to EECON2.
- 10. Write 0AAh to EECON2.
- 11. Set the WR bit. This will begin the write cycle.
- 12. The CPU will stall for the duration of the write for TIW (parameter D133A).
- 13. Re-enable interrupts.
- 14. Repeat steps 6 through 13 until all 1024 bytes are written to program memory.
- 15. Verify the memory (table read).

An example of the required code is shown in Example 7-3 on the following page.

Note: Before setting the WR bit, the Table Pointer address needs to be within the intended address range of the 64 bytes in the holding register.

EXAMPLE 7-3: WRITING TO FLASH PROGRAM MEMORY

-			
	MOLITIN		; Load TBLPTR with the base address
	MOVLW	CODE_ADDR_UPPER	
	MOVWF	TBLPTRU	; of the memory block, minus 1
	MOVLW	CODE_ADDR_HIGH	
	MOVWF	TBLPTRH	
	MOVLW	CODE_ADDR_LOW	
	MOVWF	TBLPTRL	
ERASE_BLOCK			
	BSF	EECON1, WREN	; enable write to memory
	BSF	EECON1, FREE	; enable Erase operation
	BCF	INTCON, GIE	; disable interrupts
	MOVLW	55h	
	MOVWF	EECON2	; write 55h
	MOVLW	0AAh	
	MOVWF	EECON2	; write OAAh
	BSF	EECON1, WR	; start erase (CPU stall)
	BSF	INTCON, GIE	; re-enable interrupts
	MOVLW	D'16'	
	MOVWF	WRITE_COUNTER	; Need to write 16 blocks of 64 to write
			; one erase block of 1024
RESTART_BUFFER			
	MOVLW	D'64'	
	MOVWF	COUNTER	
	MOVLW	BUFFER_ADDR_HIGH	; point to buffer
	MOVWF	FSR0H	-
	MOVLW	BUFFER_ADDR_LOW	
	MOVWF	FSROL	
FILL_BUFFER			
			; read the new data from I2C, SPI,
			; USART, etc.
WRITE_BUFFER			
	MOVLW	D'64	; number of bytes in holding register
	MOVWF	COUNTER	
WRITE_BYTE_TO_HREC	3S		
	MOVFF	POSTINC0, WREG	; get low byte of buffer data
	MOVWF	TABLAT	; present data to table latch
	TBLWT+*		; write data, perform a short write
			; to internal TBLWT holding register.
	DECFSZ	COUNTER	; loop until buffers are full
	BRA	WRITE_BYTE_TO_HREGS	
PROGRAM_MEMORY			
	BSF	EECON1, WREN	; enable write to memory
	BCF	INTCON, GIE	; disable interrupts
	MOVLW	55h	
Required	MOVWF	EECON2	; write 55h
Sequence	MOVLW	0AAh	
	MOVWF	EECON2	; write OAAh
	BSF	EECON1, WR	; start program (CPU stall)
	BSF	INTCON, GIE	; re-enable interrupts
	BCF	EECON1, WREN	; disable write to memory
		-	-
	DECFSZ	WRITE_COUNTER	; done with one write cycle
	BRA	RESTART_BUFFER	; if not done replacing the erase block
			· · · · · · · · · · · · · · · · · · ·
1			

7.5.2 FLASH PROGRAM MEMORY WRITE SEQUENCE (WORD PROGRAMMING).

The PIC18F87J90 family of devices has a feature that allows programming a single word (two bytes). This feature is enabled when the WPROG bit is set. If the memory location is already erased, the following sequence is required to enable this feature:

- 1. Load the Table Pointer register with the address of the data to be written
- 2. Write the 2 bytes into the holding registers and perform a table write

- 3. Set WPROG to enable single-word write.
- 4. Set WREN to enable write to memory.
- 5. Disable interrupts.
- 6. Write 55h to EECON2.
- 7. Write 0AAh to EECON2.
- 8. Set the WR bit. This will begin the write cycle.
- The CPU will stall for the duration of the write for Tiw (see parameter D133A).
- 10. Re-enable interrupts.

EXAMPLE 7-4: SINGLE-WORD WRITE TO FLASH PROGRAM MEMORY

	MOVLW	CODE_AI	DDR_UPPER	;	Load TBLPTR with the base address
	MOVWF	TBLPTRU	U		
	MOVLW	CODE_AI	DDR_HIGH		
	MOVWF	TBLPTRI	H		
	MOVLW	CODE_AI	DDR_LOW		
	MOVWF	TBLPTRI	L		
	MOVLW	DATA0			
	MOVWF	TABLAT			
	TBLWT*+				
	MOVLW	DATA1			
	MOVWF	TABLAT			
	TBLWT*				
PROGRAM_MEMORY					
					enable single word write
		EECON1,			enable write to memory
		INTCON,	GIE	;	disable interrupts
	MOVLW	55h			
Required		EECON2		;	write 55h
Sequence	MOVLW	0AAh			
		EECON2			write OAAh
		EECON1,			start program (CPU stall)
		INTCON,			re-enable interrupts
		EECON1,			disable single word write
	BCF	EECON1,	WREN	;	disable write to memory

7.5.3 WRITE VERIFY

Depending on the application, good programming practice may dictate that the value written to the memory should be verified against the original value. This should be used in applications where excessive writes can stress bits near the specification limit.

7.5.4 UNEXPECTED TERMINATION OF WRITE OPERATION

If a write is terminated by an unplanned event, such as loss of power or an unexpected Reset, the memory location just programmed should be verified and reprogrammed if needed. If the write operation is interrupted by a MCLR Reset or a WDT time-out Reset during normal operation, the user can check the WRERR bit and rewrite the location(s) as needed.

7.6 Flash Program Operation During Code Protection

See Section 25.6 "Program Verification and Code Protection" for details on code protection of Flash program memory.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page:
TBLPTRU	—	—	bit 21	bit 21 Program Memory Table Pointer Upper Byte (TBLPTR<20:16>)					59
TBPLTRH	Program Memory Table Pointer High Byte (TBLPTR<15:8>)						59		
TBLPTRL	Program Memory Table Pointer Low Byte (TBLPTR<7:0>)							59	
TABLAT	Program Memory Table Latch						59		
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59
EECON2	EEPROM Control Register 2 (not a physical register)							61	
EECON1			WPROG	FREE	WRERR	WREN	WR		61

TABLE 7-2: REGISTERS ASSOCIATED WITH PROGRAM FLASH MEMORY

Legend: — = unimplemented, read as '0'. Shaded cells are not used during Flash program memory access.

8.0 8 x 8 HARDWARE MULTIPLIER

8.1 Introduction

All PIC18 devices include an 8 x 8 hardware multiplier as part of the ALU. The multiplier performs an unsigned operation and yields a 16-bit result that is stored in the product register pair, PRODH:PRODL. The multiplier's operation does not affect any flags in the STATUS register.

Making multiplication a hardware operation allows it to be completed in a single instruction cycle. This has the advantages of higher computational throughput and reduced code size for multiplication algorithms and allows the PIC18 devices to be used in many applications previously reserved for digital signal processors. A comparison of various hardware and software multiply operations, along with the savings in memory and execution time, is shown in Table 8-1.

8.2 Operation

Example 8-1 shows the instruction sequence for an 8×8 unsigned multiplication. Only one instruction is required when one of the arguments is already loaded in the WREG register.

Example 8-2 shows the sequence to do an 8 x 8 signed multiplication. To account for the sign bits of the arguments, each argument's Most Significant bit (MSb) is tested and the appropriate subtractions are done.

EXAMPLE 8-1: 8 x 8 UNSIGNED MULTIPLY ROUTINE

MOVF	ARG1,	W	;
MULWF	ARG2		; ARG1 * ARG2 ->
			; PRODH:PRODL

EXAMPLE 8-2: 8 x 8 SIGNED MULTIPLY ROUTINE

MOVF	ARG1, W		
MULWF	ARG2	; ARG1 * ARG2 ->	
		; PRODH:PRODL	
BTFSC	ARG2, SB	; Test Sign Bit	
SUBWF	PRODH, F	; PRODH = PRODH	
		; – ARG1	
MOVF	ARG2, W		
BTFSC	ARG1, SB	; Test Sign Bit	
SUBWF	PRODH, F	; PRODH = PRODH	
		; – ARG2	

		Program	Cycles	Time		
Routine	Multiply Method	Memory (Words)	(Max)	@ 48 MHz	@ 10 MHz	@ 4 MHz
	Without hardware multiply	13	69	5.7 μs	27.6 μs	69 μs
8 x 8 unsigned	Hardware multiply	1	1	83.3 ns	400 ns	1 μs
8 x 8 signed	Without hardware multiply	33	91	7.5 μs	36.4 μs	91 μs
	Hardware multiply	6	6	500 ns	2.4 μs	6 μ s
16 x 16 uppigned	Without hardware multiply	21	242	20.1 μs	96.8 μs	242 μs
16 x 16 unsigned	Hardware multiply	28	28	2.3 μs	11.2 μs	28 μs
10 10 10 1	Without hardware multiply	52	254	21.6 μs	102.6 μs	254 μs
16 x 16 signed	Hardware multiply	35	40	3.3 μs	16.0 μs	40 μs

TABLE 8-1: PERFORMANCE COMPARISON FOR VARIOUS MULTIPLY OPERATIONS

Example 8-3 shows the sequence to do a 16 x 16 unsigned multiplication. Equation 8-1 shows the algorithm that is used. The 32-bit result is stored in four registers (RES3:RES0).

EQUATION 8-1: 16 x 16 UNSIGNED MULTIPLICATION ALGORITHM

RES3:RES0	=	into into into into into into into into
	=	$(ARG1H \bullet ARG2H \bullet 2^{16}) +$
		$(ARG1H \bullet ARG2L \bullet 2^8) +$
		$(ARG1L \bullet ARG2H \bullet 2^8) +$
		(ARG1L • ARG2L)
		$(AROIL \bullet ARO2L)$

EXAMPLE 8-3: 16 x 16 UNSIGNED

MULTIPLY ROUTINE

1	MOVF	ARG1L, W		
I	MULWF	ARG2L	; ARG1L * ARG	2L->
			; PRODH:PROD	L
1	MOVFF	PRODH, RES1	;	
1	MOVFF	PRODL, RESO	;	
;				
1	MOVF	ARG1H, W		
1	MULWF	ARG2H	; ARG1H * ARG	2H->
			; PRODH:PROD	L
1	MOVFF	PRODH, RES3	;	
I	MOVFF	PRODL, RES2	;	
;				
1	MOVF	ARG1L, W		
1	MULWF	ARG2H	; ARG1L * ARG	2H->
			; PRODH:PROD	L
1	MOVF	PRODL, W	;	
	ADDWF	RES1, F	; Add cross	
1	MOVF		; products	
	ADDWFC	RES2, F	;	
	CLRF	WREG	;	
	ADDWFC	RES3, F	;	
;				
1	MOVF	ARG1H, W	;	
1	MULWF	ARG2L	; ARG1H * ARG	2L->
			; PRODH:PROD	L
1	MOVF	PRODL, W	;	
	ADDWF	RES1, F	; Add cross	
1	MOVF	PRODH, W	; products	
	ADDWFC	RES2, F	;	
	CLRF	WREG	;	
	ADDWFC	RES3, F	;	
1				

Example 8-4 shows the sequence to do a 16 x 16 signed multiply. Equation 8-2 shows the algorithm used. The 32-bit result is stored in four registers (RES3:RES0). To account for the sign bits of the arguments, the MSb for each argument pair is tested and the appropriate subtractions are done.

EQUATION 8-2: 16 x 16 SIGNED MULTIPLICATION ALGORITHM

RES3:RES0=	ARG1H:ARG1L • ARG2H:ARG2L
=	$(ARG1H \bullet ARG2H \bullet 2^{16}) +$
	$(ARG1H \bullet ARG2L \bullet 2^8) +$
	$(ARG1L \bullet ARG2H \bullet 2^8) +$
	$(ARG1L \bullet ARG2L) +$
	$(-1 \bullet ARG2H < 7 > \bullet ARG1H:ARG1L \bullet 2^{16}) +$
	$(-1 \bullet ARG1H < 7 > \bullet ARG2H: ARG2L \bullet 2^{16})$

EXAMPLE 8-4: 16 x 16 SIGNED MULTIPLY ROUTINE

MOVF ARG1L, W MULWF ARG2L ; ARG1L * ARG2L -> ; PRODH:PRODL	
MOVFF PRODH, RES1 ;	
MOVFF PRODL, RESO ;	
i	
MOVF ARG1H, W	
MULWF ARG2H ; ARG1H * ARG2H ->	
; PRODH:PRODL	
MOVFF PRODH, RES3 ;	
MOVFF PRODL, RES2 ;	
;	
MOVF ARG1L, W	
MULWF ARG2H ; ARG1L * ARG2H ->	
inclui inclui inclui inclui i	
MOVF PRODL, W ;	
ADDWF RES1, F ; Add cross	
MOVF PRODH, W ; products	
ADDWFC RES2, F ;	
CLRF WREG ;	
ADDWFC RES3, F ;	
;	
MOVF ARG1H, W ;	
MULWF ARG2L ; ARG1H * ARG2L ->	
i PRODH:PRODL	
MOVF PRODL, W ;	
ADDWF RES1, F ; Add cross	
MOVF PRODH, W ; products	
ADDWFC RES2, F ;	
CLRF WREG ;	
ADDWFC RES3, F ;	
BTFSS ARG2H, 7 ; ARG2H:ARG2L neg?	,
; BTFSS ARG2H, 7 ; ARG2H:ARG2L neg? BRA SIGN_ARG1 ; no, check ARG1	
MOVF ARG1L, W ;	
SUBWF RES2 ;	
MOVF ARG1H, W ;	
SUBWFB RES3	
;	
, SIGN_ARG1	
BTFSS ARG1H, 7 ; ARG1H:ARG1L neg?	,
BRA CONT_CODE ; no, done	
MOVF ARG2L, W ;	
SUBWF RES2 ;	
MOVF ARG2H, W ;	
SUBWFB RES3	
:	
, CONT_CODE	

9.0 INTERRUPTS

Members of the PIC18F87J90 family of devices have multiple interrupt sources and an interrupt priority feature that allows most interrupt sources to be assigned a high-priority level or a low-priority level. The high-priority interrupt vector is at 0008h and the low-priority interrupt vector is at 0018h. High-priority interrupt events will interrupt any low-priority interrupts that may be in progress.

There are thirteen registers which are used to control interrupt operation. These registers are:

- RCON
- INTCON
- INTCON2
- INTCON3
- PIR1, PIR2, PIR3
- PIE1, PIE2, PIE3
- IPR1, IPR2, IPR3

It is recommended that the Microchip header files, supplied with MPLAB[®] IDE, be used for the symbolic bit names in these registers. This allows the assembler/compiler to automatically take care of the placement of these bits within the specified register.

In general, interrupt sources have three bits to control their operation. They are:

- Flag bit to indicate that an interrupt event occurred
- Enable bit that allows program execution to branch to the interrupt vector address when the flag bit is set
- **Priority bit** to select high priority or low priority

The interrupt priority feature is enabled by setting the IPEN bit (RCON<7>). When interrupt priority is enabled, there are two bits which enable interrupts globally. Setting the GIEH bit (INTCON<7>) enables all interrupts that have the priority bit set (high priority). Setting the GIEL bit (INTCON<6>) enables all interrupts that have the priority bit cleared (low priority). When the interrupt flag, enable bit and appropriate global interrupt enable bit are set, the interrupt will vector immediately to address, 0008h or 0018h, depending on the priority bit setting. Individual interrupts can be disabled through their corresponding enable bits.

When the IPEN bit is cleared (default state), the interrupt priority feature is disabled and interrupts are compatible with PIC[®] mid-range devices. In Compatibility mode, the interrupt priority bits for each source have no effect. INTCON<6> is the PEIE bit which enables/disables all peripheral interrupt sources. INTCON<7> is the GIE bit which enables/disables all interrupt sources. All interrupts branch to address, 0008h, in Compatibility mode.

When an interrupt is responded to, the global interrupt enable bit is cleared to disable further interrupts. If the IPEN bit is cleared, this is the GIE bit. If interrupt priority levels are used, this will be either the GIEH or GIEL bit. High-priority interrupt sources can interrupt a low-priority interrupt. Low-priority interrupts are not processed while high-priority interrupts are in progress.

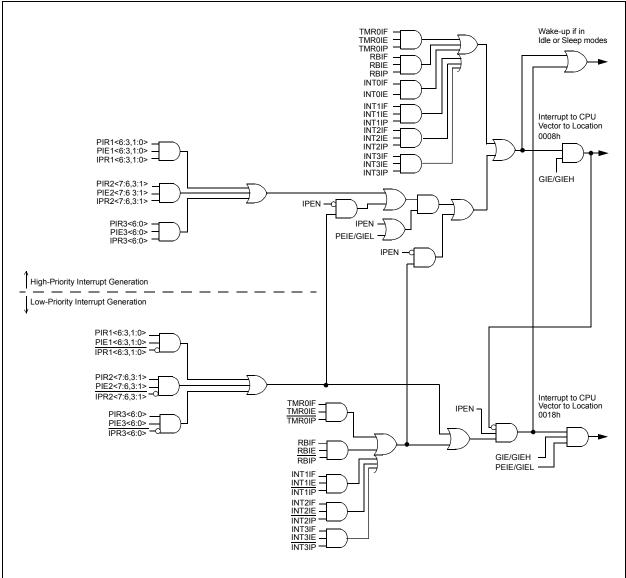
The return address is pushed onto the stack and the PC is loaded with the interrupt vector address (0008h or 0018h). Once in the Interrupt Service Routine, the source(s) of the interrupt can be determined by polling the interrupt flag bits. The interrupt flag bits must be cleared in software before re-enabling interrupts to avoid recursive interrupts.

The "return from interrupt" instruction, RETFIE, exits the interrupt routine and sets the GIE bit (GIEH or GIEL if priority levels are used) which re-enables interrupts.

For external interrupt events, such as the INTx pins or the PORTB input change interrupt, the interrupt latency will be three to four instruction cycles. The exact latency is the same for one or two-cycle instructions. Individual interrupt flag bits are set regardless of the status of their corresponding enable bit or the GIE bit.

Note: Do not use the MOVFF instruction to modify any of the Interrupt Control registers while any interrupt is enabled. Doing so may cause erratic microcontroller behavior.





9.1 INTCON Registers

The INTCON registers are readable and writable registers which contain various enable, priority and flag bits.

Note: Interrupt flag bits are set when an interrupt condition occurs regardless of the state of its corresponding enable bit or the global interrupt enable bit. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt. This feature allows for software polling.

REGISTER 9-1: INTCON: INTERRUPT CONTROL REGISTER

GIE/GIEH PEIE/GIEL TMR0IE INT0IE RBIE TMR0IF INT0IF RBIF ⁽¹⁾ bit 7 bit 0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-x
bit 7 bit 0	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF ⁽¹⁾
	bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	GIE/GIEH: Global Interrupt Enable bit
	When IPEN = 0:
	1 = Enables all unmasked interrupts
	0 = Disables all interrupts
	<u>When IPEN = 1:</u> 1 = Enables all high-priority interrupts
	0 = Disables all interrupts
bit 6	PEIE/GIEL: Peripheral Interrupt Enable bit
	When IPEN = 0:
	1 = Enables all unmasked peripheral interrupts
	0 = Disables all peripheral interrupts
	When IPEN = 1:
	1 = Enables all low-priority peripheral interrupts
6:4 C	0 = Disables all low-priority peripheral interrupts
bit 5	TMR0IE: TMR0 Overflow Interrupt Enable bit
	 Enables the TMR0 overflow interrupt Disables the TMR0 overflow interrupt
bit 4	INTOIE: INTO External Interrupt Enable bit
	1 = Enables the INT0 external interrupt
	0 = Disables the INT0 external interrupt
bit 3	RBIE: RB Port Change Interrupt Enable bit
	1 = Enables the RB port change interrupt
	0 = Disables the RB port change interrupt
bit 2	TMR0IF: TMR0 Overflow Interrupt Flag bit
	1 = TMR0 register has overflowed (must be cleared in software)
	0 = TMR0 register did not overflow
bit 1	INT0IF: INT0 External Interrupt Flag bit
	 1 = The INT0 external interrupt occurred (must be cleared in software) 0 = The INT0 external interrupt did not occur
bit 0	RBIF: RB Port Change Interrupt Flag bit ⁽¹⁾
	1 = At least one of the RB<7:4> pins changed state (must be cleared in software)
	0 = None of the RB<7:4> pins have changed state
Noto 1	A mismatch condition will continue to set this hit. Reading DOPTR, and then waiting one additional instruction
Note 1:	A mismatch condition will continue to set this bit. Reading PORTB, and then waiting one additional instruction

cycle, will end the mismatch condition and allow the bit to be cleared.

bit 7 Legend: R = Readable bit -n = Value at POR bit 7 RB 1 = 0 = bit 6 INT 1 =	PU: PORT All PORT PORTB p EDG0: Ext	INTEDG1 W = Writable t '1' = Bit is set B Pull-up Enab B pull-ups are o ull-ups are ena ternal Interrupt	le bit disabled	U = Unimplem '0' = Bit is clea		INT3IP d as '0' x = Bit is unkn	RBIP bit	
Legend: R = Readable bit -n = Value at POR bit 7 RB 1 = 0 = bit 6 INT 1 =	PU: PORT All PORT PORTB p EDG0: Ext	'1' = Bit is set B Pull-up Enab B pull-ups are o ull-ups are ena	le bit disabled	-				
R = Readable bit -n = Value at POR bit 7 RB 1 = 0 = bit 6 INT 1 =	PU: PORT All PORT PORTB p EDG0: Ext	'1' = Bit is set B Pull-up Enab B pull-ups are o ull-ups are ena	le bit disabled	-			own	
bit 7 RB 1 = 0 = bit 6 INT 1 =	PU: PORT All PORT PORTB p EDG0: Ext	'1' = Bit is set B Pull-up Enab B pull-ups are o ull-ups are ena	le bit disabled	-			own	
bit 7 RB 1 = 0 = bit 6 INT 1 =	PU: PORT All PORT PORTB p EDG0: Ext	B Pull-up Enab B pull-ups are o ull-ups are ena	disabled	ʻ0' = Bit is clea	ared	x = Bit is unkn	own	
1 = 0 = bit 6 INT 1 =	All PORT PORTB p EDG0: Ext	B pull-ups are o ull-ups are ena	disabled					
1 = 0 = bit 6 INT 1 =	All PORT PORTB p EDG0: Ext	B pull-ups are o ull-ups are ena	disabled					
bit 6 INT 1 =	EDG0: Ex	•	ماتينا المستحيية المسالية					
1 =		ternal Interrupt	iblea by individ	dual port latch v	alues			
	Interrupt of		0 Edge Select	t bit				
0 =		on rising edge						
		on falling edge						
		ternal Interrupt	1 Edge Select	t bit				
	 Interrupt on rising edge Interrupt on falling edge 							
		ternal Interrupt	2 Edge Select	t bit				
		on rising edge	z Luge Ocicei					
		on falling edge						
bit 3 INT	EDG3: Ext	ternal Interrupt	3 Edge Select	t bit				
1 =	Interrupt of	on rising edge	-					
0 =	Interrupt of	on falling edge						
bit 2 TM	ROIP: TMF	R0 Overflow Inte	errupt Priority	bit				
	High prior							
	Low prior	•						
		External Interru	ipt Priority bit					
	High prior	•						
	-	t Change Interi	runt Priority bit					
	High prior	•	upt i nonty bit					
	Low prior							
	·	-						

REGISTER 9-2: INTCON2: INTERRUPT CONTROL REGISTER 2

Note: Interrupt flag bits are set when an interrupt condition occurs regardless of the state of its corresponding enable bit or the global interrupt enable bit. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt. This feature allows for software polling.

REGISTER 9-3: INTCON3: INTERRUPT CONTROL REGISTER 3

R/W-1	R/W-1	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
INT2IP	INT1IP	INT3IE	INT2IE	INT1IE	INT3IF	INT2IF	INT1IF
bit 7	•	·		•			bit (
Legend:							
R = Reada	able bit	W = Writable	bit	U = Unimplen	nented bit, rea	d as '0'	
-n = Value	at POR	'1' = Bit is set		'0' = Bit is clea		x = Bit is unkr	nown
bit 7	INT2IP: INT2	2 External Interr	upt Priority bit				
	1 = High prid0 = Low prid	•					
bit 6	INT1IP: INT1	External Interr	upt Priority bit				
	1 = High prid0 = Low prid	,					
bit 5	INT3IE: INT3	B External Interr	upt Enable bit				
		the INT3 extern the INT3 extern					
bit 4	INT2IE: INT2	2 External Interr	upt Enable bit				
		the INT2 extern					
L H 0		the INT2 exter					
bit 3		External Interr	•				
		the INT1 exter					
bit 2	INT3IF: INT3	External Interr	upt Flag bit				
		3 external interi 3 external interi		must be cleared	d in software)		
bit 1	INT2IF: INT2	External Interr	upt Flag bit				
		2 external interi 2 external interi	•	must be cleared	d in software)		
bit 0	INT1IF: INT1	External Interr	upt Flag bit				
		1 external intern 1 external intern		must be cleared cur	d in software)		
Note:	Interrupt flag bits	s are set when	an interrupt co	ondition occurs	regardless of	the state of its	corresponding
	enable bit or the are clear prior to	global interrupt	enable bit. Us	er software sho	uld ensure the	e appropriate int	

9.2 PIR Registers

The PIR registers contain the individual flag bits for the peripheral interrupts. Due to the number of peripheral interrupt sources, there are three Peripheral Interrupt Request (Flag) registers (PIR1, PIR2, PIR3).

- Note 1: Interrupt flag bits are set when an interrupt condition occurs regardless of the state of its corresponding enable bit or the Global Interrupt Enable bit, GIE (INTCON<7>).
 - 2: User software should ensure the appropriate interrupt flag bits are cleared prior to enabling an interrupt and after servicing that interrupt.

REGISTER 9-4: PIR1: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 1

U-0	R/W-0	R-0	R-0	R/W-0	U-0	R/W-0	R/W-0
—	ADIF	RC1IF	TX1IF	SSPIF	—	TMR2IF	TMR1IF
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	Unimplemented: Read as '0'
bit 6	ADIF: A/D Converter Interrupt Flag bit
	1 = An A/D conversion completed (must be cleared in software)0 = The A/D conversion is not complete
bit 5	RC1IF: EUSART Receive Interrupt Flag bit
	 1 = The EUSART receive buffer, RCREG1, is full (cleared when RCREG1 is read) 0 = The EUSART receive buffer is empty
bit 4	TX1IF: EUSART Transmit Interrupt Flag bit
	 1 = The EUSART transmit buffer, TXREG1, is empty (cleared when TXREG1 is written) 0 = The EUSART transmit buffer is full
bit 3	SSPIF: Master Synchronous Serial Port Interrupt Flag bit
	1 = The transmission/reception is complete (must be cleared in software)0 = Waiting to transmit/receive
bit 2	Unimplemented: Read as '0'
bit 1	TMR2IF: TMR2 to PR2 Match Interrupt Flag bit
	1 = TMR2 to PR2 match occurred (must be cleared in software)
	0 = No TMR2 to PR2 match occurred
bit 0	TMR1IF: TMR1 Overflow Interrupt Flag bit
	1 = TMR1 register overflowed (must be cleared in software)
	0 = TMR1 register did not overflow

REGISTER 9-5: PIR2: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 2

R/W-0	R/W-0	U-0	U-0	R/W-0	R/W-0	R/W-0	U-0
OSCFIF	CMIF	—		BCLIF	LVDIF	TMR3IF	—
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	OSCFIF: Oscillator Fail Interrupt Flag bit 1 = Device oscillator failed, clock input has changed to INTOSC (must be cleared in software) 0 = Device clock operating
bit 6	CMIF: Comparator Interrupt Flag bit
	 1 = Comparator input has changed (must be cleared in software) 0 = Comparator input has not changed
bit 5-4	Unimplemented: Read as '0'
bit 3	BCLIF: Bus Collision Interrupt Flag bit
	 1 = A bus collision occurred (must be cleared in software) 0 = No bus collision occurred
bit 2	LVDIF: Low-Voltage Detect Interrupt Flag bit
	 1 = A low-voltage condition occurred (must be cleared in software) 0 = The device voltage is above the regulator's low-voltage trip point
bit 1	TMR3IF: TMR3 Overflow Interrupt Flag bit
	 1 = TMR3 register overflowed (must be cleared in software) 0 = TMR3 register did not overflow
bit 0	Unimplemented: Read as '0'

U-0	R/W-0	R-0	R-0	R/W-0	R/W-0	R/W-0	R/W-0
	LCDIF	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF
bit 7							bi
_egend:							
R = Reada	ble bit	W = Writable	bit	U = Unimplen	nented bit, read	d as '0'	
-n = Value at POR		'1' = Bit is set		'0' = Bit is cleared		x = Bit is unknown	
bit 7	Unimplemer	ented: Read as '0'					
bit 6	LCDIF: LCD Interrupt Flag bit (valid when Type-B waveform with Non-Static mode is selected)						
	 1 = LCD data of all COMs is output (must be cleared in software) 0 = LCD data of all COMs is not yet output 						
bit 5	RC2IF: AUSART Receive Interrupt Flag bit						
	 1 = The AUSART receive buffer, RCREG2, is full (cleared when RCREG2 is read) 0 = The AUSART receive buffer is empty 						
bit 4	TX2IF: AUSART Transmit Interrupt Flag bit						
	1 = The AUSART transmit buffer, TXREG2, is empty (cleared when TXREG2 is written)						
	0 = The AUSART transmit buffer is full						
bit 3	CTMUIF: CTMU Interrupt Flag bit						
	1 = CTMU interrupt occured (must be cleared in software)						
	0 = No CTMU interrupt occured						
bit 2	CCP2IF: CCP2 Interrupt Flag bit						
	<u>Capture mode:</u> 1 = A TMR1/TMR3 register capture occurred (must be cleared in software)						
	0 = No TMR1/TMR3 register capture occurred						
	Compare mode:						
	1 = A TMR1/TMR3 register compare match occurred (must be cleared in software)						
	0 = No TMR1/TMR3 register compare match occurred						
	<u>PWM mode:</u> Unused in this mode.						
bit 1	CCP1IF: CCP1 Interrupt Flag bit						
	Capture mode:						
	1 = A TMR1/TMR3 register capture occurred (must be cleared in software)						
	0 = No TMR1/TMR3 register capture occurred						
	Compare mode:						
	 1 = A TMR1/TMR3 register compare match occurred (must be cleared in software) 0 = No TMR1/TMR3 register compare match occurred 						
	<u>PWM mode:</u>						
	Unused in this mode.						
bit 0	RTCCIF: RTCC Interrupt Flag bit						
	1 = RTCC interrupt occured (must be cleared in software)						
	0 = No RTC	C interrupt occu	red				

9.3 PIE Registers

The PIE registers contain the individual enable bits for the peripheral interrupts. Due to the number of peripheral interrupt sources, there are three Peripheral Interrupt Enable registers (PIE1, PIE2, PIE3). When IPEN = 0, the PEIE bit must be set to enable any of these peripheral interrupts.

REGISTER 9-7: PIE1: PERIPHERAL INTERRUPT ENABLE REGISTER 1

U-0	R/W-0	R/W-0	R/W-0	R/W-0	U-0	R/W-0	R/W-0
—	ADIE	RC1IE	TX1IE	SSPIE	—	TMR2IE	TMR1IE
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	Unimplemented: Read as '0'
bit 6	ADIE: A/D Converter Interrupt Enable bit
	1 = Enables the A/D interrupt
	0 = Disables the A/D interrupt
bit 5	RC1IE: EUSART Receive Interrupt Enable bit
	1 = Enables the EUSART receive interrupt
	0 = Disables the EUSART receive interrupt
bit 4	TX1IE: EUSART Transmit Interrupt Enable bit
	1 = Enables the EUSART transmit interrupt
	0 = Disables the EUSART transmit interrupt
bit 3	SSPIE: Master Synchronous Serial Port Interrupt Enable bit
bit 3	SSPIE: Master Synchronous Serial Port Interrupt Enable bit 1 = Enables the MSSP interrupt
bit 3	
bit 3 bit 2	1 = Enables the MSSP interrupt
	1 = Enables the MSSP interrupt0 = Disables the MSSP interrupt
bit 2	 1 = Enables the MSSP interrupt 0 = Disables the MSSP interrupt Unimplemented: Read as '0'
bit 2	 1 = Enables the MSSP interrupt 0 = Disables the MSSP interrupt Unimplemented: Read as '0' TMR2IE: TMR2 to PR2 Match Interrupt Enable bit
bit 2	 1 = Enables the MSSP interrupt 0 = Disables the MSSP interrupt Unimplemented: Read as '0' TMR2IE: TMR2 to PR2 Match Interrupt Enable bit 1 = Enables the TMR2 to PR2 match interrupt
bit 2 bit 1	 1 = Enables the MSSP interrupt 0 = Disables the MSSP interrupt Unimplemented: Read as '0' TMR2IE: TMR2 to PR2 Match Interrupt Enable bit 1 = Enables the TMR2 to PR2 match interrupt 0 = Disables the TMR2 to PR2 match interrupt
bit 2 bit 1	 1 = Enables the MSSP interrupt 0 = Disables the MSSP interrupt Unimplemented: Read as '0' TMR2IE: TMR2 to PR2 Match Interrupt Enable bit 1 = Enables the TMR2 to PR2 match interrupt 0 = Disables the TMR2 to PR2 match interrupt TMR1IE: TMR1 Overflow Interrupt Enable bit

R/W-0	R/W-0	U-0	U-0	R/W-0	R/W-0	R/W-0	U-0
OSCFIE	CMIE	—	—	BCLIE	LVDIE	TMR3IE	—
bit 7							bit 0
Legend:							
R = Readable	e bit	W = Writable	bit	U = Unimplem	nented bit, read	d as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkno	own
bit 7	OSCFIE: Osc	illator Fail Inter	rupt Enable bi	t			
	1 = Enabled						
	0 = Disabled						
bit 6	•	arator Interrupt	Enable bit				
	1 = Enabled 0 = Disabled						
bit 5-4		ted. Dood oo '	<u>.</u> '				
	-	ted: Read as '					
bit 3		Collision Interru	pt Enable bit				
	1 = Enabled 0 = Disabled						
bit 2							
DIL Z	LVDIE: Low-Voltage Detect Interrupt Enable bit 1 = Enabled						
	0 = Disabled						
bit 1	TMR3IE: TMR3 Overflow Interrupt Enable bit						
	1 = Enabled						
	0 = Disabled						
bit 0	Unimplemen	ted: Read as '	כ'				

REGISTER 9-8: PIE2: PERIPHERAL INTERRUPT ENABLE REGISTER 2

REGISTER 9-9: PIE3: PERIPHERAL INTERRUPT ENABLE REGISTER 3

U-0	R/W-0	R-0	R-0	R/W-0	R/W-0	R/W-0	R/W-0
_	LCDIE	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE
oit 7		•	•		•		bit 0
Legend:							
R = Readabl	e bit	W = Writable	bit	U = Unimplem	nented bit, rea	d as '0'	
-n = Value at	t POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown
bit 7	Unimpleme	nted: Read as '	כ'				
bit 6	LCDIE: LCD	Interrupt Enable	e bit (valid whe	n Type-B wavefo	orm with Non-	Static mode is so	elected)
	1 = Enabled						
	0 = Disable	-					
oit 5		SART Receive In	terrupt Enable	e bit			
	1 = Enabled						
	0 = Disable						
bit 4		ART Transmit Ir	iterrupt Enable	e bit			
	1 = Enabled	-					
L:1 0	0 = Disable						
bit 3		FMU Interrupt Er	hable bit				
	1 = Enabled 0 = Disabled	-					
bit 2		CP2 Interrupt Ena	ahla hit				
	1 = Enabled	-					
	0 = Disable						
bit 1	CCP1IE: CC	P1 Interrupt Ena	able bit				
		the CCP1 interr					
		the CCP1 inter					
bit 0	RTCCIE: RT	CC Interrupt En	able bit				
	1 = Enabled	•					
	0 = Disable	d					

9.4 IPR Registers

The IPR registers contain the individual priority bits for the peripheral interrupts. Due to the number of peripheral interrupt sources, there are three Peripheral Interrupt Priority registers (IPR1, IPR2, IPR3). Using the priority bits requires that the Interrupt Priority Enable (IPEN) bit be set.

REGISTER 9-10: IPR1: PERIPHERAL INTERRUPT PRIORITY REGISTER 1

U-0	R/W-1	R/W-1	R/W-1	R/W-1	U-0	R/W-1	R/W-1
_	ADIP	RC1IP	TX1IP	SSPIP	—	TMR2IP	TMR1IP
bit 7				·			bit 0
Legend:	ala hit		L:1		nonted hit was	d aa '0'	
R = Readat		W = Writable	DIL	•	nented bit, rea		
-n = Value a	at POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	IOWN
bit 7	Unimpleme	nted: Read as ')'				
bit 6	ADIP: A/D C	onverter Interru	pt Priority bit				
	1 = High prie	ority					
	0 = Low pric	ority					
bit 5	RC1IP: EUS	ART Receive In	terrupt Priority	' bit			
	1 = High prie						
	0 = Low pric	ority					
bit 4	TX1IP: EUS	ART Transmit In	terrupt Priority	/ bit			
	1 = High prie	•					
	0 = Low pric	ority					
bit 3	SSPIP: Mast	ter Synchronous	Serial Port In	terrupt Priority	bit		
	1 = High prie						
	0 = Low pric	•					
bit 2	Unimpleme	nted: Read as ')'				
bit 1	TMR2IP: TM	IR2 to PR2 Mate	h Interrupt Pri	iority bit			
	1 = High price	•					
	0 = Low pric	•					
bit 0		IR1 Overflow Int	errupt Priority	bit			
	1 = High price						
	0 = Low pric	лцу					

REGISTER 9-11: IPR2: PERIPHERAL INTERRUPT PRIORITY REGISTER 2

R/W-1	R/W-1	U-0	U-0	R/W-1	R/W-1	R/W-1	U-0
OSCFIP	CMIP	—		BCLIP	LVDIP	TMR3IP	—
bit 7							bit 0
Logondu							
Legend:			- :4			-l (O)	
R = Readat		W = Writable I	DIC	U = Unimplem			
-n = Value a	at POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkno	own
bit 7		cillator Fail Inter	rupt Priority bil	t			
	1 = High price	•					
	0 = Low prio	5	.				
bit 6	•	arator Interrupt	Priority bit				
	1 = High price	•					
	0 = Low prio	•					
bit 5-4	Unimplemer	nted: Read as '0)'				
bit 3	BCLIP: Bus	Collision Interru	ot Priority bit				
	1 = High prio	,					
	0 = Low prio	rity					
bit 2	LVDIP: Low-	Voltage Detect I	nterrupt Priorit	y bit			
	1 = High prio	ority					
	0 = Low prio	rity					
bit 1	TMR3IP: TM	R3 Overflow Int	errupt Priority	bit			
	1 = High prio	ority					
	0 = Low prio	rity					
bit 0	Unimplemer	nted: Read as 'o)'				

U-0	R/W-1	R-1	R-1	R/W-1	R/W-1	R/W-1	R/W-1
—	LCDIP	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP
bit 7							bit C
Legend:							
R = Readab	ole bit	W = Writable	bit	U = Unimplem	nented bit, read	d as '0'	
-n = Value a	at POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown
bit 7	-	ted: Read as '					
bit 6			bit (valid whe	n Type-B wavefo	orm with Non-S	Static mode is se	elected)
	1 = High prio 0 = Low prior						
bit 5	•	-	iority Flog bit				
DIL 5	1 = High prio	ART Receive Pi	IONLY FIAG DI				
	0 = Low prior	•					
bit 4	•	RT Transmit Ir	terrupt Priority	y bit			
	1 = High prio			, ,			
	0 = Low prior	ity					
bit 3	CTMUIP: CTM	MU Interrupt Pr	iority bit				
	1 = High prio						
	0 = Low prior	5					
bit		P2 Interrupt Pri	ority bit				
	1 = High prio 0 = Low prior	•					
bit	•	P1 Interrupt Pri	ority hit				
bit	1 = High prio	•	Sincy Sic				
	0 = Low prior						
bit 0	RTCCIP: RTC	CC Interrupt Pri	ority bit				
	1 = High prio	rity					
	0 = Low prior	ity					

REGISTER 9-12: IPR3: PERIPHERAL INTERRUPT PRIORITY REGISTER 3

9.5 RCON Register

The RCON register contains bits used to determine the cause of the last Reset or wake-up from Idle or Sleep modes. RCON also contains the bit that enables interrupt priorities (IPEN).

REGISTER 9-13: RCON: RESET CONTROL REGISTER

R/W-0	U-0	R/W-1	R/W-1	R-1	R-1	R/W-0	R/W-0
IPEN	_	CM	RI	TO	PD	POR	BOR
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	t, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	 IPEN: Interrupt Priority Enable bit 1 = Enable priority levels on interrupts 0 = Disable priority levels on interrupts (PIC16CXXX Compatibility mode)
bit 6	Unimplemented: Read as '0'
bit 5	CM: Configuration Mismatch Flag bit
	 1 = A Configuration Mismatch Reset has not occurred 0 = A Configuration Mismatch Reset has occurred (must be subsequently set in software)
bit 4	RI: RESET Instruction Flag bit
	For details of bit operation, see Register 5-1.
bit 3	TO: Watchdog Timer Time-out Flag bit
	For details of bit operation, see Register 5-1.
bit 2	PD: Power-Down Detection Flag bit
	For details of bit operation, see Register 5-1.
bit 1	POR: Power-on Reset Status bit
	For details of bit operation, see Register 5-1.
bit 0	BOR: Brown-out Reset Status bit
	For details of bit operation, see Register 5-1.

9.6 INTx Pin Interrupts

External interrupts on the RB0/INT0, RB1/INT1, RB2/INT2 and RB3/INT3 pins are edge-triggered. If the corresponding INTEDGx bit in the INTCON2 register is set (= 1), the interrupt is triggered by a rising edge; if the bit is clear, the trigger is on the falling edge. When a valid edge appears on the RBx/INTx pin, the corresponding flag bit, INTxIF, is set. This interrupt can be disabled by clearing the corresponding enable bit, INTxIE. Flag bit, INTxIF, must be cleared in software in the Interrupt Service Routine before re-enabling the interrupt.

All external interrupts (INT0, INT1, INT2 and INT3) can wake-up the processor from the power-managed modes if bit, INTxIE, was set prior to going into the power-managed modes. If the Global Interrupt Enable bit, GIE, is set, the processor will branch to the interrupt vector following wake-up.

Interrupt priority for INT1, INT2 and INT3 is determined by the value contained in the Interrupt Priority bits, INT1IP (INTCON3<6>), INT2IP (INTCON3<7>) and INT3IP (INTCON2<1>). There is no priority bit associated with INT0. It is always a high-priority interrupt source.

9.7 TMR0 Interrupt

In 8-bit mode (which is the default), an overflow in the TMR0 register (FFh \rightarrow 00h) will set flag bit, TMR0IF. In 16-bit mode, an overflow in the TMR0H:TMR0L register pair (FFFFh \rightarrow 0000h) will set TMR0IF. The interrupt can be enabled/disabled by setting/clearing enable bit, TMR0IE (INTCON<5>). Interrupt priority for Timer0 is determined by the value contained in the interrupt priority bit, TMR0IP (INTCON2<2>). See Section 11.0 "Timer0 Module" for further details on the Timer0 module.

9.8 **PORTB Interrupt-on-Change**

An input change on PORTB<7:4> sets flag bit, RBIF (INTCON<0>). The interrupt can be enabled/disabled by setting/clearing enable bit, RBIE (INTCON<3>). Interrupt priority for PORTB interrupt-on-change is determined by the value contained in the interrupt priority bit, RBIP (INTCON2<0>).

9.9 Context Saving During Interrupts

During interrupts, the return PC address is saved on the stack. Additionally, the WREG, STATUS and BSR registers are saved on the Fast Return Stack. If a fast return from interrupt is not used (see **Section 6.3 "Data Memory Organization"**), the user may need to save the WREG, STATUS and BSR registers on entry to the Interrupt Service Routine (ISR). Depending on the user's application, other registers may also need to be saved. Example 9-1 saves and restores the WREG, STATUS and BSR registers during an Interrupt Service Routine.

EXAMPLE 9-1:	SAVING STATUS, WREG AND BSR REGISTERS IN RAM
--------------	--

MOVWF MOVFF MOVFF ;	W_TEMP STATUS, STATUS_TEMP BSR, BSR_TEMP	; W_TEMP is in virtual bank ; STATUS_TEMP located anywhere ; BSR_TMEP located anywhere
; USER	ISR CODE	
, MOVFF MOVF MOVFF	BSR_TEMP, BSR W_TEMP, W STATUS_TEMP, STATUS	; Restore BSR ; Restore WREG ; Restore STATUS

10.0 I/O PORTS

Depending on the device selected and features enabled, there are up to nine ports available. Some pins of the I/O ports are multiplexed with an alternate function from the peripheral features on the device. In general, when a peripheral is enabled, that pin may not be used as a general purpose I/O pin.

Each port has three memory mapped registers for its operation:

- TRIS register (Data Direction register)
- PORT register (reads the levels on the pins of the device)
- LAT register (Output Latch register)

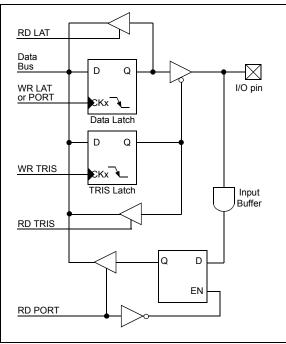
Reading the PORT register reads the current status of the pins, whereas writing to the PORT register writes to the Output Latch (LAT) register.

Setting a TRIS bit (= 1) makes the corresponding port pin an input (i.e., puts the corresponding output driver in a high-impedance mode). Clearing a TRIS bit (= 0) makes the corresponding port pin an output (i.e., puts the contents of the corresponding LAT bit on the selected pin).

The Output Latch (LAT register) is useful for read-modify-write operations on the value that the I/O pins are driving. Read-modify-write operations on the LAT register read and write the latched output value for the PORT register.

A simplified model of a generic I/O port, without the interfaces to other peripherals, is shown in Figure 10-1.

FIGURE 10-1: GENERIC I/O PORT OPERATION



10.1 I/O Port Pin Capabilities

When developing an application, the capabilities of the port pins must be considered. Outputs on some pins have higher output drive strength than others. Similarly, some pins can tolerate higher than VDD input levels.

10.1.1 INPUT PINS AND VOLTAGE CONSIDERATIONS

The voltage tolerance of pins used as device inputs is dependent on the pin's input function. Pins that are used as digital only inputs are able to handle DC voltages up to 5.5V, a level typical for digital logic circuits. In contrast, pins that also have analog input functions of any kind can only tolerate voltages up to VDD. Voltage excursions beyond VDD on these pins should be avoided.

Table 10-1 summarizes the input voltage capabilities. Refer to **Section 28.0 "Electrical Characteristics"** for more details.

TABLE 10-1: INPUT VOLTAGE TOLERANCE

PORT or Pin	Tolerated Input	Description	
PORTA<7:0>		Only VDD input levels	
PORTC<1:0>	Voo	tolerated.	
PORTF<7:1>	VUU		
PORTG<3:2>			
PORTB<7:0>		Tolerates input	
PORTC<7:2>		levels above VDD,	
PORTD<7:0>		useful for most standard logic.	
PORTE<7:3>	5.5V	otariaara logio.	
PORTG<4,1>			
PORTH<7:0> ⁽¹⁾			
PORTJ<7:0> ⁽¹⁾			

Note 1: Not available on PIC18F6XJ90 devices.

10.1.2 PIN OUTPUT DRIVE

When used as digital I/O, the output pin drive strengths vary for groups of pins intended to meet the needs for a variety of applications. In general, there are three classes of output pins in terms of drive capability.

PORTB and PORTC, as well as PORTA<7:6>, are designed to drive higher current loads, such as LEDs. PORTD, PORTE and PORTJ can also drive LEDs but only those with smaller current requirements. PORTF, PORTG and PORTH, along with PORTA<5:0>, have the lowest drive level but are capable of driving normal digital circuit loads with a high input impedance. Regardless of which port it is located on, all output pins in LCD Segment or Common mode have sufficient output to directly drive a display.

Table 10-2 summarizes the output capabilities of the ports. Refer to the "Absolute Maximum Ratings" in Section 28.0 "Electrical Characteristics" for more details.

TABLE 10-2:OUTPUT DRIVE LEVELS FOR
VARIOUS PORTS

Low	Medium	High
PORTA<5:0>	PORTD	PORTA<7:6>
PORTF	PORTE	PORTB
PORTG	PORTJ ⁽¹⁾	PORTC
PORTH ⁽¹⁾		

Note 1: Not available on PIC18F6XJ90 devices.

10.1.3 PULL-UP CONFIGURATION

Four of the I/O ports (PORTB, PORTD, PORTE and PORTJ) implement configurable weak pull-ups on all pins. These are internal pull-ups that allow floating digital input signals to be pulled to a consistent level without the use of external resistors.

The pull-ups are enabled with a single bit for each of the ports: RBPU (INTCON2<7>) for PORTB, and RDPU, REPU and PJPU (PORTG<7:5>) for the other ports.

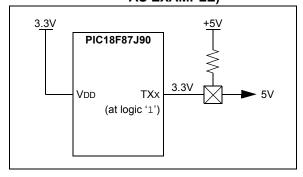
10.1.4 OPEN-DRAIN OUTPUTS

The output pins for several peripherals are also equipped with a configurable, open-drain output option. This allows the peripherals to communicate with external digital logic, operating at a higher voltage level, without the use of level translators.

The open-drain option is implemented on port pins specifically associated with the data and clock outputs of the USARTs, the MSSP module (in SPI mode) and the CCP modules. This option is selectively enabled by setting the open-drain control bit for the corresponding module in TRISG and LATG. Their configuration is discussed in more detail in Section 10.4 "PORTC, TRISC and LATC Registers", Section 10.6 "PORTE, TRISE and LATE Registers" and Section 10.8 "PORTG, TRISG and LATG Registers".

When the open-drain option is required, the output pin must also be tied through an external pull-up resistor, provided by the user to a higher voltage level, up to 5V (Figure 10-2). When a digital logic high signal is output, it is pulled up to the higher voltage level.

FIGURE 10-2: USING THE OPEN-DRAIN OUTPUT (USART SHOWN AS EXAMPLE)



10.2 PORTA, TRISA and LATA Registers

PORTA is an 8-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISA and LATA.

RA4/T0CKI is a Schmitt Trigger input. All other PORTA pins have TTL input levels and full CMOS output drivers.

The RA4 pin is multiplexed with the Timer0 clock input and one of the LCD segment drives. RA5 and RA<3:0> are multiplexed with analog inputs for the A/D Converter.

The operation of the analog inputs as A/D Converter inputs is selected by clearing or setting the PCFG<3:0> control bits in the ADCON1 register. The corresponding TRISA bits control the direction of these pins, even when they are being used as analog inputs. The user must ensure the bits in the TRISA register are maintained set when using them as analog inputs.

Note:	RA5 and RA<3:0> are configured as
	analog inputs on any Reset and are read
	as '0'. RA4 is configured as a digital input.

OSC2/CLKO/RA6 and OSC1/CLKI/RA7 normally serve as the external circuit connections for the external (primary) oscillator circuit (HS Oscillator modes), or the external clock input and output (EC Oscillator modes). In these cases, RA6 and RA7 are not available as digital I/O and their corresponding TRIS and LAT bits are read as '0'. When the device is configured to use INTOSC or INTRC as the default oscillator mode (FOSC2 Configuration bit is '0'), RA6 and RA7 are automatically configured as digital I/O. The oscillator and clock in/clock out functions are disabled.

RA1, RA4 and RA5 are multiplexed with LCD segment drives, controlled by bits in the LCDSE1 and LCDSE2 registers. I/O port functionality is only available when the LCD segments are disabled.

EXAMPLE 10-1: INITIALIZING PORTA

CLRF	PORTA	; Initialize PORTA by
		; clearing output latches
CLRF	LATA	; Alternate method to
		; clear output data latches
MOVLW	07h	; Configure A/D
MOVWF	ADCON1	; for digital inputs
MOVLW	0BFh	; Value used to initialize
		; data direction
MOVWF	TRISA	; Set RA<7, 5:0> as inputs,
		; RA<6> as output

TABLE 10-3.		TRIS I/O		I/O	Description		
Pin Name	Function	Setting	I/O	Туре	Description		
RA0/AN0	RA0	0	0	DIG	LATA<0> data output; not affected by analog input.		
		1	Ι	TTL	PORTA<0> data input; disabled when analog input enabled.		
	AN0	1	I	ANA	A/D Input Channel 0. Default input configuration on POR; does not affect digital output.		
RA1/AN1/SEG18	RA1 0 O DIG LATA<1> data output; not affected by analog input.				LATA<1> data output; not affected by analog input.		
		1	I	TTL	PORTA<1> data input; disabled when analog input enabled.		
	AN1	1	I	ANA	A/D Input Channel 1. Default input configuration on POR; does not affect digital output.		
	SEG18	x	0	ANA	LCD Segment 18 output; disables all other pin functions.		
RA2/AN2/VREF-	RA2	0	0	DIG	LATA<2> data output; not affected by analog input.		
		1	I	TTL	PORTA<2> data input; disabled when analog functions enabled.		
	AN2	1	I	ANA	A/D Input Channel 2. Default input configuration on POR.		
	VREF-	1	I	ANA	A/D and comparator low reference voltage input.		
RA3/AN3/VREF+	RA3	0	0	DIG	LATA<3> data output; not affected by analog input.		
		1	I	TTL	PORTA<3> data input; disabled when analog input enabled.		
	AN3	1	I	ANA	A/D Input Channel 3. Default input configuration on POR.		
	VREF+	1	I	ANA	A/D and comparator high reference voltage input.		
RA4/T0CKI/	RA4	0	0	DIG	LATA<4> data output.		
SEG14		1	I	ST	PORTA<4> data input. Default configuration on POR.		
	TOCKI	x	Ι	ST	Timer0 clock input.		
	SEG14	x	0	ANA	LCD Segment 14 output; disables all other pin functions.		
RA5/AN4/SEG15	RA5	0	0	DIG	LATA<5> data output; not affected by analog input.		
		1	I	TTL	PORTA<5> data input; disabled when analog input enabled.		
	AN4	1	I	ANA	A/D Input Channel 4. Default configuration on POR.		
	SEG15	x	0	ANA	LCD Segment 15 output; disables all other pin functions.		
OSC2/CLKO/RA6	OSC2	x	0	ANA	Main oscillator feedback output connection (HS and HSPLL modes).		
	CLKO	x	0	DIG	System cycle clock output (Fosc/4) (EC and ECPLL modes).		
	RA6	0	0	DIG	LATA<6> data output; disabled when FOSC2 Configuration bit is set.		
		1	I	TTL	PORTA<6> data input; disabled when FOSC2 Configuration bit is set		
OSC1/CLKI/RA7	OSC1	x	I	ANA	Main oscillator input connection (HS and HSPLL modes).		
	CLKI	x	I	ANA	Main external clock source input (EC and ECPLL modes).		
	RA7	0	0	DIG	LATA<7> data output; disabled when FOSC2 Configuration bit is set.		
		1	I	TTL	PORTA<7> data input; disabled when FOSC2 Configuration bit is set		

TABLE 10-3: PORTA FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, TTL = TTL Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
PORTA	RA7 ⁽¹⁾	RA6 ⁽¹⁾	RA5	RA4	RA3	RA2	RA1	RA0	63
LATA	LATA7 ⁽¹⁾	LATA6 ⁽¹⁾	LATA5	LATA4	LATA3	LATA2	LATA1	LATA0	62
TRISA	TRISA7 ⁽¹⁾	TRISA6 ⁽¹⁾	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	62
ADCON1	TRIGSEL	_	VCFG1	VCFG0	PCFG3	PCFG2	PCFG1	PCFG0	61
LCDSE1	SE15	SE14	SE13	SE12	SE11	SE10	SE09	SE08	61
LCDSE2	SE23	SE22	SE21	SE20	SE19	SE18	SE17	SE16	61

Legend: — = unimplemented, read as '0'. Shaded cells are not used by PORTA.

Note 1: These bits are enabled depending on the oscillator mode selected. When not enabled as PORTA pins, they are disabled and read as 'x'.

10.3 PORTB, TRISB and LATB Registers

PORTB is an 8-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISB and LATB. All pins on PORTB are digital only and tolerate voltages up to 5.5V.

EXAMPLE 10-2: INITIALIZING PORTB

CLRF	PORTB	; Initialize PORTB by ; clearing output
67. D. D.		; data latches
CLRF	LATB	; Alternate method ; to clear output
		; data latches
MOVLW	OCFh	; Value used to
		; initialize data
		; direction
MOVWF	TRISB	; Set RB<3:0> as inputs
		; RB<5:4> as outputs
		; RB<7:6> as inputs

Each of the PORTB pins has a weak internal pull-up. A single control bit can turn <u>on all</u> the pull-ups. This is performed by clearing bit, RBPU (INTCON2<7>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on a Power-on Reset.

Four of the PORTB pins (RB<7:4>) have an interrupt-on-change feature. Only pins configured as inputs can cause this interrupt to occur (i.e., any RB<7:4> pin configured as an output is excluded from the interrupt-on-change comparison). The input pins (of RB<7:4>) are compared with the old value latched on the last read of PORTB. The "mismatch" outputs of RB<7:4> are ORed together to generate the RB Port Change Interrupt with Flag bit, RBIF (INTCON<0>).

This interrupt can wake the device from power-managed modes. The user, in the Interrupt Service Routine, can clear the interrupt in the following manner:

- a) Any read or write of PORTB (except with the MOVFF (ANY), PORTB instruction). This will end the mismatch condition.
- b) Wait one instruction cycle (such as executing a NOP instruction).
- c) Clear flag bit, RBIF.

A mismatch condition will continue to set flag bit, RBIF. Reading PORTB will end the mismatch condition and allow flag bit, RBIF, to be cleared after one TCY delay.

The interrupt-on-change feature is recommended for wake-up on key depression operation and operations where PORTB is only used for the interrupt-on-change feature. Polling of PORTB is not recommended while using the interrupt-on-change feature.

RB<3:2> are multiplexed as CTMU edge inputs.

RB<5:0> are also multiplexed with LCD segment drives, controlled by bits in the LCDSE1 and LCDSE3 registers. I/O port functionality is only available when the LCD segments are disabled.

Pin Name	Function	TRIS Setting	I/O	I/О Туре	Description			
RB0/INT0/SEG30	RB0	0	0	DIG	LATB<0> data output.			
		1	I	TTL	PORTB<0> data input; weak pull-up when RBPU bit is cleared.			
	INT0	1	I	ST	External Interrupt 0 input.			
	SEG30	x	0	ANA	LCD Segment 30 output; disables all other pin functions.			
RB1/INT1/SEG8	RB1	0	0	DIG	LATB<1> data output.			
		1	I	TTL	PORTB<1> data input; weak pull-up when RBPU bit is cleared.			
	INT1	1	I	ST	External Interrupt 1 input.			
	SEG8	x	0	ANA	LCD Segment 8 output; disables all other pin functions.			
RB2/INT2/SEG9/	RB2	0	0	DIG	LATB<2> data output.			
CTED1		1	Ι	TTL	PORTB<2> data input; weak pull-up when RBPU bit is cleared.			
	INT2	1	I	ST	External Interrupt 2 input.			
	SEG9	x	0	ANA	LCD Segment 9 output; disables all other pin functions.			
	CTED1	x	Ι	ST	CTMU Edge 1 input.			
RB3/INT3/SEG10/	RB3	0	0	DIG	LATB<3> data output.			
CTED2		1	I	TTL	PORTB<3> data input; weak pull-up when RBPU bit is cleared.			
	INT3	1	I	ST	External Interrupt 3 input.			
	SEG10	x	0	ANA	LCD Segment 10 output; disables all other pin functions.			
	CTED2	x	-	ST	CTMU Edge 2 input.			
RB4/KBI0/SEG11	RB4	0	0	DIG	LATB<4> data output.			
		1	Ι	TTL	PORTB<4> data input; weak pull-up when RBPU bit is cleared.			
	KBI0	1	I	TTL	Interrupt-on-pin change.			
	SEG11	x	0	ANA	LCD Segment 11 output; disables all other pin functions.			
RB5/KBI1/SEG29	RB5	0	0	DIG	LATB<5> data output.			
		1	Ι	TTL	PORTB<5> data input; weak pull-up when RBPU bit is cleared.			
	KBI1	1	I	TTL	Interrupt-on-pin change.			
	SEG29	x	0	ANA	LCD Segment 29 output; disables all other pin functions.			
RB6/KBI2/PGC	RB6	0	0	DIG	LATB<6> data output.			
		1	I	TTL	PORTB<6> data input; weak pull-up when RBPU bit is cleared.			
	KBI2	1	Ι	TTL	Interrupt-on-pin change.			
	PGC	x	I	ST	Serial execution (ICSP™) clock input for ICSP and ICD operation			
RB7/KBI3/PGD	RB7	0	0	DIG	LATB<7> data output.			
		1	I	TTL	PORTB<7> data input; weak pull-up when RBPU bit is cleared.			
	KBI3	1	I	TTL	Interrupt-on-pin change.			
	PGD	x	0	DIG	Serial execution data output for ICSP and ICD operation.			
		x	I	ST	Serial execution data input for ICSP and ICD operation.			

TABLE 10-5: PORTB FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, TTL = TTL Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	63
LATB	LATB7	LATB6	LATB5	LATB4	LATB3	LATB2	LATB1	LATB0	62
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	62
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59
INTCON2	RBPU	INTEDG0	INTEDG1	INTEDG2	INTEDG3	TMR0IP	INT3IP	RBIP	59
INTCON3	INT2IP	INT1IP	INT3IE	INT2IE	INT1IE	INT3IF	INT2IF	INT1IF	59
LCDSE1	SE15	SE14	SE13	SE12	SE11	SE10	SE09	SE08	61
LCDSE3	SE31	SE30	SE29	SE28	SE27	SE26	SE25	SE24	61

TABLE 10-6: SUMMARY OF REGISTERS ASSOCIATED WITH PORTB

Legend: Shaded cells are not used by PORTB.

10.4 PORTC, TRISC and LATC Registers

PORTC is an 8-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISC and LATC. Only PORTC pins, RC2 through RC7, are digital only pins and can tolerate input voltages up to 5.5V.

PORTC is multiplexed with CCP, MSSP and EUSART peripheral functions (Table 10-7). The pins have Schmitt Trigger input buffers. The pins for CCP, SPI and EUSART are also configurable for open-drain output whenever these functions are active. Open-drain configuration is selected by setting the SPIOD, CCPxOD and U1OD control bits (TRISG<7:5> and LATG<6>, respectively).

RC1 is normally configured as the default peripheral pin for the CCP2 module. Assignment of CCP2 is controlled by Configuration bit, CCP2MX (default state, CCP2MX = 1).

When enabling peripheral functions, care should be taken in defining TRIS bits for each PORTC pin. Some peripherals override the TRIS bit to make a pin an output, while other peripherals override the TRIS bit to make a pin an input. The user should refer to the corresponding peripheral section for the correct TRIS bit settings.

Note: These pins are configured as digital inputs on any device Reset.

The contents of the TRISC register are affected by peripheral overrides. Reading TRISC always returns the current contents, even though a peripheral device may be overriding one or more of the pins.

RC<7:1> pins are multiplexed with LCD segment drives, controlled by bits in the LCDSE1, LCDSE2, LCDSE3 and LCDSE4 registers. I/O port functionality is only available when the LCD segments are disabled.

EXAMPLE 10-3:	INITIALIZING PORTC

CLRF PORTC	; Initialize PORTC by ; clearing output
	; data latches
CLRF LATC	; Alternate method
	; to clear output
	; data latches
MOVLW 0CFh	; Value used to
	; initialize data
	; direction
MOVWF TRISC	; Set RC<3:0> as inputs
	; RC<5:4> as outputs
	; RC<7:6> as inputs

Pin Name	Function	TRIS Setting	I/O	l/O Type	Description		
RC0/T1OSO/	RC0	0	0	DIG	LATC<0> data output.		
T13CKI		1	-	ST	PORTC<0> data input.		
	T10S0	x	0	ANA	Timer1 oscillator output; enabled when Timer1 oscillator enabled. Disables digital I/O and LCD segment driver.		
	T13CKI	1	Ι	ST	Timer1/Timer3 counter input.		
RC1/T1OSI/	RC1	0	0	DIG	LATC<1> data output.		
CCP2/SEG32		1	Ι	ST	PORTC<1> data input.		
	T1OSI	x	Ι	ANA	Timer1 oscillator input.		
	CCP2 ⁽¹⁾	0	0	DIG	CCP2 compare/PWM output.		
		1	Ι	ST	CCP2 capture input.		
	SEG32	x	0	ANA	LCD Segment 32 output; disables all other pin functions.		
RC2/CCP1/	RC2	0	0	DIG	LATC<2> data output.		
SEG13		1	Ι	ST	PORTC<2> data input.		
	CCP1	0	0	DIG	CCP1 compare/PWM output; takes priority over port data.		
		1	Ι	ST	CCP1 capture input.		
	SEG13	x	0	ANA	LCD Segment 13 output; disables all other pin functions.		
RC3/SCK/SCL/	RC3	0	0	DIG	LATC<3> data output.		
SEG17		1	-	ST	PORTC<3> data input.		
	SCK	0	0	DIG	SPI clock output (MSSP module); takes priority over port data.		
		1	Ι	ST	SPI clock input (MSSP module).		
	SCL	0	0	DIG	I ² C [™] clock output (MSSP module); takes priority over port data.		
		1	Ι	I2C	I ² C clock input (MSSP module); input type depends on module setting.		
	SEG17	x	0	ANA	LCD Segment 17 output; disables all other pin functions.		
RC4/SDI/SDA/	RC4	0	0	DIG	LATC<4> data output.		
SEG16		1	Ι	ST	PORTC<4> data input.		
	SDI		Ι	ST	SPI data input (MSSP module).		
	SDA	1	0	DIG	I ² C data output (MSSP module); takes priority over port data.		
		1	Ι	I2C	I ² C data input (MSSP module); input type depends on module setting.		
	SEG16	x	0	ANA	LCD Segment 16 output; disables all other pin functions.		
RC5/SDO/	RC5	0	0	DIG	LATC<5> data output.		
SEG12		1	Ι	ST	PORTC<5> data input.		
	SDO	0	0	DIG	SPI data output (MSSP module).		
	SEG12	x	0	ANA	LCD Segment 12 output; disables all other pin functions.		
RC6/TX1/CK1/	RC6	0	0	DIG	LATC<6> data output.		
SEG27		1	Ι	ST	PORTC<6> data input.		
	TX1	1	0	DIG	Synchronous serial data output (EUSART module); takes priority over port data.		
	CK1	1	0	DIG	Synchronous serial data input (EUSART module); user must configure as an input		
		1	Ι	ST	Synchronous serial clock input (EUSART module).		
	SEG27	x	0	ANA	LCD Segment 27 output; disables all other pin functions.		
RC7/RX1/DT1/	RC7	0	0	DIG	LATC<7> data output.		
SEG28		1	Ι	ST	PORTC<7> data input.		
	RX1	1	I	ST	Asynchronous serial receive data input (EUSART module).		
	DT1	1	0	DIG	Synchronous serial data output (EUSART module); takes priority over port data.		
		1	Ι	ST	Synchronous serial data input (EUSART module); user must configure as an input		
	SEG28	x	0	ANA	LCD Segment 28 output; disables all other pin functions.		

TABLE 10-7: PORTC FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, TTL = TTL Buffer Input, $I2C = I^2C/SMBus$ Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Note 1: Default assignment for CCP2 when CCP2MX Configuration bit is set.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
PORTC	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0	63
LATC	LATC7	LATBC6	LATC5	LATCB4	LATC3	LATC2	LATC1	LATC0	62
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	62
LATG	U2OD	U10D	_	LATG4	LATG3	LATG2	LATG1	LATG0	62
TRISG	SPIOD	CCP2OD	CCP10D	TRISG4	TRISG3	TRISG2	TRISG1	TRISG0	62
LCDSE1	SE15	SE14	SE13	SE12	SE11	SE10	SE09	SE08	61
LCDSE2	SE23	SE22	SE21	SE20	SE19	SE18	SE17	SE16	61
LCDSE3	SE31	SE30	SE29	SE28	SE27	SE26	SE25	SE24	61
LCDSE4	SE39 ⁽¹⁾	SE38 ⁽¹⁾	SE37 ⁽¹⁾	SE36 ⁽¹⁾	SE35 ⁽¹⁾	SE34 ⁽¹⁾	SE33 ⁽¹⁾	SE32	61

TABLE 10-8: SUMMARY OF REGISTERS ASSOCIATED WITH PORTC

Legend: Shaded cells are not used by PORTC.

Note 1: Unimplemented on PIC18F6XJ90 devices, read as '0'.

10.5 PORTD, TRISD and LATD Registers

PORTD is an 8-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISD and LATD. All pins on PORTD are digital only and tolerate voltages up to 5.5V.

All pins on PORTD are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output.

Note:	These pins are configured as digital inputs
	on any device Reset.

Each of the PORTD pins has a weak internal pull-up. A single control bit can turn off all the pull-ups. This is performed by clearing bit, RDPU (PORTG<7>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on all device Resets.

All of the PORTD pins are multiplexed with LCD segment drives, controlled by bits in the LCDSE0 register. RD0 is multiplexed with the CTMU pulse generator output.

I/O port functionality is only available when the LCD segments are disabled.

EXAMPLE 10-4:	INITIALIZING PORTD

CLRF	PORTD	; Initialize PORTD by ; clearing output
67.5.5		; data latches
CLRF	LATD	; Alternate method ; to clear output
		; data latches
MOVLW	OCFh	; Value used to
		; initialize data ; direction
MOVWF	TRISD	; Set RD<3:0> as inputs
		; RD<5:4> as outputs
		; RD<7:6> as inputs

TABLE 10-9: PORTD FUNCTIONS						
Pin Name	Function	TRIS Setting	I/O	l/O Type	Description	
RD0/SEG0/	RD0	0	0	DIG	LATD<0> data output.	
CTPLS		1	I	ST	PORTD<0> data input.	
	SEG0	x	0	ANA	LCD Segment 0 output; disables all other pin functions.	
	CTPLS	x	0	DIG	CTMU pulse generator output	
RD1/SEG1	RD1	0	0	DIG	LATD<1> data output.	
		1	I	ST	PORTD<1> data input.	
	SEG1	x	0	ANA	LCD Segment 1 output; disables all other pin functions.	
RD2/SEG2	RD2	0	0	DIG	LATD<2> data output.	
		1	Ι	ST	PORTD<2> data input.	
	SEG2	x	0	ANA	LCD Segment 2 output; disables all other pin functions.	
RD3/SEG3	RD3	0	0	DIG	LATD<3> data output.	
		1	Ι	ST	PORTD<3> data input.	
	SEG3	x	0	ANA	LCD Segment 3 output; disables all other pin functions.	
RD4/SEG4	RD4	0	0	DIG	LATD<4> data output.	
		1	Ι	ST	PORTD<4> data input.	
	SEG4	x	0	ANA	LCD Segment 4 output; disables all other pin functions.	
RD5/SEG5	RD5	0	0	DIG	LATD<5> data output.	
		1	Ι	ST	PORTD<5> data input.	
	SEG5	x	0	ANA	LCD Segment 5 output; disables all other pin functions.	
RD6/SEG6	RD6	0	0	DIG	LATD<6> data output.	
		1	Ι	ST	PORTD<6> data input.	
	SEG6	x	0	ANA	LCD Segment 6 output; disables all other pin functions.	
RD7/SEG7	RD7	0	0	DIG	LATD<7> data output.	
		1	Ι	ST	PORTD<7> data input.	
	SEG7	x	I	ANA	LCD Segment 7 output; disables all other pin functions.	

TABLE 10-9:PORTD FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
PORTD	RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0	63
LATD	LATD7	LATD6	LATD5	LATD4	LATD3	LATD2	LATD1	LATD0	62
TRISD	TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0	62
PORTG	RDPU	REPU	RJPU ⁽¹⁾	RG4	RG3	RG2	RG1	RG0	62
LCDSE0	SE07	SE06	SE05	SE04	SE03	SE02	SE01	SE00	61

Legend: Shaded cells are not used by PORTD.

Note 1: Unimplemented on PIC18F6XJ90 devices, read as '0'.

10.6 PORTE, TRISE and LATE Registers

PORTE is a 7-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISE and LATE. All pins on PORTE are digital only and tolerate voltages up to 5.5V.

All pins on PORTE are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output. The RE7 pin is also configurable for open-drain output when CCP2 is active on this pin. Open-drain configuration is selected by setting the CCP2OD control bit (TRISG<6>)

Note:	These pins are configured as digital inputs	
	on any device Reset.	

Each of the PORTE pins has a weak internal pull-up. A single control bit can turn off all the pull-ups. This is performed by clearing bit, REPU (PORTG<6>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on any device Reset.

Pins, RE<6:3>, are multiplexed with the LCD common drives. I/O port functions are only available on those PORTE pins depending on which commons are active. The configuration is determined by the LMUX<1:0> control bits (LCDCON<1:0>). The availability is summarized in Table 10-11.

TABLE 10-11: PORTE PINS AVAILABLE IN DIFFERENT LCD DRIVE CONFIGURATIONS

LCDCON <1:0>	Active LCD Commons	PORTE Available for I/O
00	COM0	RE6, RE5, RE4
01	COM0, COM1	RE6, RE5
10	COM0, COM1 and COM2	RE6
11	All (COM0 through COM3)	None

Pins, RE1 and RE0, are multiplexed with the functions of LCDBIAS2 and LCDBIAS1. When LCD bias generation is required (i.e., any application where the device is connected to an external LCD), these pins cannot be used as digital I/O.

Note:	The pin corresponding to RE2 of other
	PIC18F parts has the function of
	LCDBIAS3 in this device. It cannot be used
	as digital I/O.

RE7 is multiplexed with the LCD segment drive (SEG31) controlled by the LCDSE3<7> bit. I/O port function is only available when the segment is disabled.

RE7 can also be configured as the alternate peripheral pin for the CCP2 module. This is done by clearing the CCP2MX Configuration bit.

EXAMPLE 10-5: INITIALIZING PORTE

CLRF	PORTE	; Initialize PORTE by ; clearing output
CLRF	LATE	; data latches ; Alternate method ; to clear output
MOVLW	03h	; data latches ; Value used to ; initialize data
MOVWF	TRISE	; direction ; Set RE<1:0> as inputs ; RE<7:2> as outputs

Pin Name	Function	TRIS Setting	I/O	I/O Туре	Description
RE0/LCDBIAS1	RE0	0	0	DIG	LATE<0> data output.
		1	Ι	ST	PORTE<0> data input.
	LCDBIAS1	—	Ι	ANA	LCD module bias voltage input.
RE1/LCDBIAS2	RE1	0	0	DIG	LATE<1> data output.
		1		ST	PORTE<1> data input.
	LCDBIAS2	_		ANA	LCD module bias voltage input.
RE3/COM0	RE3	0	0	DIG	LATE<3> data output.
		1		ST	PORTE<3> data input.
	COM0	x	0	ANA	LCD Common 0 output; disables all other outputs.
RE4/COM1	RE4	0	0	DIG	LATE<4> data output.
		1		ST	PORTE<4> data input.
	COM1	x	0	ANA	LCD Common 1 output; disables all other outputs.
RE5/COM2	RE5	0	0	DIG	LATE<5> data output.
		1		ST	PORTE<5> data input.
	COM2	x	0	ANA	LCD Common 2 output; disables all other outputs.
RE6/COM3	RE6	0	0	DIG	LATE<6> data output.
		1		ST	PORTE<6> data input.
	COM3	x	0	ANA	LCD Common 3 output; disables all other outputs.
RE7/CCP2/	RE7	0	0	DIG	LATE<7> data output.
SEG31		1	Ι	ST	PORTE<7> data input.
	CCP2 ⁽¹⁾	0	0	DIG	CCP2 compare/PWM output; takes priority over port data.
		1	Ι	ST	CCP2 capture input.
	SEG31	х	0	ANA	Segment 31 analog output for LCD; disables digital output.

TABLE 10-12: PORTE FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Note 1: Alternate assignment for CCP2 when CCP2MX Configuration bit is cleared.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
PORTE	RE7	RE6	RE5	RE4	RE3	_	RE1	RE0	63
LATE	LATE7	LATE6	LATE5	LATE4	LATE3	—	LATE1	LATE0	62
TRISE	TRISE7	TRISE6	TRISE5	TRISE4	TRISE3	—	TRISE1	TRISE0	62
PORTG	RDPU	REPU	RJPU ⁽¹⁾	RG4	RG3	RG2	RG1	RG0	62
TRISG	SPIOD	CCP2OD	CCP10D	TRISG4	TRISG3	TRISG2	TRISG1	TRISG0	62
LCDCON	LCDEN	SLPEN	WERR	_	CS1	CS0	LMUX1	LMUX0	61
LCDSE3	SE31	SE30	SE29	SE28	SE27	SE26	SE25	SE24	61

Legend: Shaded cells are not used by PORTE.

Note 1: Unimplemented on PIC18F6XJ90 devices, read as '0'.

10.7 PORTF, LATF and TRISF Registers

PORTF is a 7-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISF and LATF. All pins on PORTF are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output.

PORTF is multiplexed with analog peripheral functions, as well as LCD segments. Pins, RF1 through RF6, may be used as comparator inputs or outputs by setting the appropriate bits in the CMCON register. To use RF<6:3> as digital inputs, it is also necessary to turn off the comparators.

- **Note 1:** On device Resets, pins, RF<6:1>, are configured as analog inputs and are read as '0'.
 - 2: To configure PORTF as digital I/O, turn off the comparators and set the ADCON1 value.

PORTF is also multiplexed with LCD segment drives controlled by bits in the LCDSE2 and LCDSE3 registers. I/O port functions are only available when the segments are disabled.

EXAMPLE 10-6: INITIALIZING PORTF

Initialize PORTF by clearing output data latches
Alternate method to clear output data latches
Turn off comparators
Set PORTF as digital I/O
Value used to
initialize data
direction
Set RF3:RF1 as inputs
RF5:RF4 as outputs
RF7:RF6 as inputs

TABLE 10-14: Pin Name	Function	TRIS Setting	1/0	1/O	Description					
		-	-	Туре	·					
RF1/AN6/C2OUT/ SEG19	RF1	0	0	DIG	LATF<1> data output; not affected by analog input.					
02019		1		ST	PORTF<1> data input; disabled when analog input is enabled.					
	AN6	1		ANA	A/D Input Channel 6. Default configuration on POR.					
	C2OUT	0	0	DIG	Comparator 2 output; takes priority over port data.					
	SEG19	х	0	ANA	LCD Segment 19 output; disables all other pin functions.					
					LATF<2> data output; not affected by analog input.					
SEG20		1	I	ST	PORTF<2> data input; disabled when analog input is enabled.					
	AN7	1	Ι	ANA	A/D Input Channel 7. Default configuration on POR.					
	C10UT	0	0	DIG	Comparator 1 output; takes priority over port data.					
	SEG20	х	0	ANA	A LCD Segment 20 output; disables all other pin functions.					
RF3/AN8/SEG21/	RF3	0	0	DIG	LATF<3> data output; not affected by analog input.					
C2INB		1	-	ST	PORTF<3> data input; disabled when analog input is enabled.					
	AN8	1	I	ANA	A/D Input Channel 8 and Comparator C2+ input. Default input configuration on POR; not affected by analog output.					
	SEG21	х	0	ANA	LCD Segment 21 output; disables all other pin functions.					
	C2INB	1	Ι	ANA	Comparator 2 Input B.					
RF4/AN9/SEG22/	RF4	0	0	DIG	LATF<4> data output; not affected by analog input.					
C2INA		1	-	ST	PORTF<4> data input; disabled when analog input is enabled.					
	AN9	1	Ι	ANA	A/D Input Channel 9 and Comparator C2- input. Default input configuration on POR; does not affect digital output.					
	SEG22	х	0	ANA	LCD Segment 22 output; disables all other pin functions.					
	C2INA	1	Ι	ANA	Comparator 2 Input A.					
RF5/AN10/CVREF/ SEG23/C1INB	RF5	0	0	DIG	LATF<5> data output; not affected by analog input. Disabled when CVREF output is enabled.					
		1	I	ST	PORTF<5> data input; disabled when analog input is enabled. Disabled when CVREF output is enabled.					
	AN10	1	Ι	ANA	A/D Input Channel 10 and Comparator C1+ input. Default input configuration on POR.					
	CVREF	х	0	ANA	Comparator voltage reference output. Enabling this feature disables digital I/O.					
	SEG23	х	0	ANA	LCD Segment 23 output; disables all other pin functions.					
	C1INB	1	-	ANA	Comparator 1 Input B.					
RF6/AN11/SEG24/	RF6	0	0	DIG	LATF<6> data output; not affected by analog input.					
C1INA		1	Ι	ST	PORTF<6> data input; disabled when analog input is enabled.					
	AN11	1	-	ANA	A/D Input Channel 11 and Comparator C1- input. Default input configuration on POR; does not affect digital output.					
	SEG24	x	0	ANA	LCD Segment 24 output; disables all other pin functions.					
	C1INA	1	Ι	ANA	Comparator 1 Input A.					
RF7/AN5/SS/	RF7	0	0	DIG	LATF<7> data output; not affected by analog input.					
SEG25		1		ST	PORTF<7> data input; disabled when analog input is enabled.					
	AN5	1		ANA	A/D Input Channel 5. Default configuration on POR.					
		1		TTL	Slave select input for MSSP module.					
			-							
Legend: $O = Out$		EG25 x O ANA LCD Segment 25 output; disables all other pin functions. L = Input ANA = Apalog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input								

TABLE 10-14: PORTF FUNCTIONS

O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, Legend:

TTL = TTL Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
PORTF	RF7	RF6	RF5	RF4	RF3	RF2	RF1	_	62
LATF	LATF7	LATF6	LATF5	LATF4	LATF3	LATF2	LATF1	_	62
TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1	—	62
ADCON1	TRIGSEL	_	VCFG1	VCFG0	PCFG3	PCFG2	PCFG1	PCFG0	61
CMCON	C2OUT	C10UT	C2INV	C1INV	CIS	CM2	CM1	CM0	61
CVRCON	CVREN	CVROE	CVRR	CVRSS	CVR3	CVR2	CVR1	CVR0	61
LCDSE2	SE23	SE22	SE21	SE20	SE19	SE18	SE17	SE16	61
LCDSE3	SE31	SE30	SE29	SE28	SE27	SE26	SE25	SE24	61

TABLE 10-15: SUMMARY OF REGISTERS ASSOCIATED WITH PORTF

Legend: — = unimplemented, read as '0'. Shaded cells are not used by PORTF.

10.8 PORTG, TRISG and LATG Registers

PORTG is a 5-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISG and LATG. All pins on PORTG are digital only and tolerate voltages up to 5.5V.

PORTG is multiplexed with both AUSART and LCD functions (Table 10-16). When operating as I/O, all PORTG pins have Schmitt Trigger input buffers. The RG1 pin is also configurable for open-drain output when the AUSART is active. Open-drain configuration is selected by setting the U2OD control bit (LATG<7>).

RG4 is multiplexed with LCD segment drives controlled by bits in the LCDSE2 register and as the RTCC pin. The I/O port function is only available when the segments are disabled.

RG3 and RG2 are multiplexed with the VLCAP pins for the LCD charge pump and RG0 is multiplexed with the LCDBIAS0 bias voltage input. When these pins are used for LCD bias generation, the I/O and other functions are unavailable.

When enabling peripheral functions, care should be taken in defining TRIS bits for each PORTG pin. Some peripherals override the TRIS bit to make a pin an output, while other peripherals override the TRIS bit to make a pin an input. The user should refer to the corresponding peripheral section for the correct TRIS bit settings. The pin override value is not loaded into the TRIS register. This allows read-modify-write of the TRIS register without concern due to peripheral overrides. Although the port itself is only five bits wide, the PORTG<7:5> bits are still implemented to control the weak pull-ups on the I/O ports associated with PORTD, PORTE and PORTJ. Clearing these bits enables the respective port pull-ups. All pull-ups are disabled by default on all device Resets.

Most of the corresponding TRISG and LATG bits are implemented as open-drain control bits for CCP1, CCP2 and SPI (TRISG<7:5>), and the USARTs (LATG<7:6>). Setting these bits configures the output pin for the corresponding peripheral for open-drain operation. LATG<5> is not implemented.

EXAMPLE 10-7:	INITIALIZING PORTG

CLRF	PORTG	; Initialize PORTG by ; clearing output
CLRF	LATG	<pre>; data latches ; Alternate method ; to clear output ; data latches</pre>
MOVLW	04h	<pre>; Value used to ; initialize data ; direction</pre>
MOVWF	TRISG	; Set RG1:RG0 as outputs ; RG2 as input ; RG4:RG3 as inputs

Pin Name	Function	TRIS Setting	I/O	l/O Type	Description			
RG0/LCDBIAS0	RG0	0	0	DIG	LATG<0> data output.			
			Ι	ST	PORTG<0> data input.			
	LCDBIAS0	х	Ι	ANA	LCD module bias voltage input.			
RG1/TX2/CK2	RG1	0	0	DIG	LATG<1> data output.			
		1	Ι	ST	PORTG<1> data input.			
	TX2	1	0	DIG	Synchronous serial data output (AUSART module); takes priority over port data.			
	CK2	1	0	DIG	Synchronous serial data input (AUSART module); user must configure as an input.			
		1	Ι	ST	Synchronous serial clock input (AUSART module).			
RG2/RX2/DT2/	RG2	0	0	DIG	LATG<2> data output.			
VLCAP1		1	I	ST	PORTG<2> data input.			
	RX2	1	Ι	ST	Asynchronous serial receive data input (AUSART module).			
	DT2	1	0	DIG	Synchronous serial data output (AUSART module); takes priority over port data.			
		1	Ι	ST	Synchronous serial data input (AUSART module); user must configure as an input.			
	VLCAP1	х	Ι	ANA	LCD charge pump capacitor input.			
RG3/VLCAP2	RG3	0	0	DIG	LATG<3> data output.			
		1	Ι	ST	PORTG<3> data input.			
	VLCAP2	х	Ι	ANA	LCD charge pump capacitor input.			
RG4/SEG26/	RG4	0	0	DIG	LATG<4> data output.			
RTCC		1	Ι	ST	PORTG<4> data input.			
	SEG26	х	0	ANA	LCD Segment 26 output; disables all other pin functions.			
	RTCC	х	0	DIG	RTCC output.			

TABLE 10-16: PORTG FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

TABLE 10-17:	SUMMARY OF REGISTERS ASSOCIATED WITH PORTG
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Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
PORTG	RDPU	REPU	RJPU ⁽¹⁾	RG4	RG3	RG2	RG1	RG0	62
LATG	U2OD	U10D	_	LATG4	LATG3	LATG2	LATG1	LATG0	62
TRISG	SPIOD	CCP2OD	CCP10D	TRISG4	TRISG3	TRISG2	TRISG1	TRISG0	62
LCDSE3	SE31	SE30	SE29	SE28	SE27	SE26	SE25	SE24	61

Legend: — = unimplemented, read as '0'. Shaded cells are not used by PORTG.

Note 1: Unimplemented on PIC18F6XJ90 devices, read as '0'.

10.9 PORTH, LATH and TRISH Registers

Note:	PORTH is available only on PIC18F8XJ90
	devices.

PORTH is an 8-bit wide, bidirectional I/O port. The corresponding Data Direction and Output Latch registers are TRISH and LATH. All pins are digital only and tolerate voltages up to 5.5V.

All pins on PORTH are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output.

All PORTH pins are multiplexed with LCD segment drives controlled by the LCDSE5 register. I/O port functions are only available when the segments are disabled.

EXAMP	LE 10-8:	INITIALIZING PORTH
CLRF	PORTH	; Initialize PORTH by
		; clearing output
		; data latches
CLRF	LATH	; Alternate method
		; to clear output
		; data latches
MOVLW	0Fh	; Configure PORTH as
MOVWF	ADCON1	; digital I/O
MOVLW	0CFh	; Value used to
		; initialize data
		; direction
MOVWF	TRISH	; Set RH3:RH0 as inputs
		; RH5:RH4 as outputs
		; RH7:RH6 as inputs

Pin Name	Function	TRIS Setting	I/O	I/O Type	Description
RH0/SEG47	RH0	0	0	DIG	LATH<0> data output.
	1	Ι	ST	PORTH<0> data input.	
	SEG47	x	0	ANA	LCD Segment 47 output; disables all other pin functions.
RH1/SEG46	RH1	0	0	DIG	LATH<1> data output.
		1	I	ST	PORTH<1> data input.
	SEG46	x	0	ANA	LCD Segment 46 output; disables all other pin functions.
RH2/SEG45	RH2	0	0	DIG	LATH<2> data output.
		1	Ι	ST	PORTH<2> data input.
	SEG45	x	0	ANA	LCD Segment 45 output; disables all other pin functions.
RH3/SEG44 RH3		0	0	DIG	LATH<3> data output.
		1	Ι	ST	PORTH<3> data input.
	SEG44	x	0	ANA	LCD Segment 44 output; disables all other pin functions.
RH4/SEG40	RH4	0	0	DIG	LATH<4> data output.
		1	Ι	ST	PORTH<4> data input.
	SEG40	x	0	ANA	LCD Segment 40 output; disables all other pin functions.
RH5/SEG41 RH5 0		0	DIG	LATH<5> data output.	
		1	Ι	ST	PORTH<5> data input.
	SEG41	x	0	ANA	LCD Segment 41 output; disables all other pin functions.
RH6/SEG42	RH6	0	0	DIG	LATH<6> data output.
		1	Ι	ST	PORTH<6> data input.
	SEG42	x	0	ANA	LCD Segment 42 output; disables all other pin functions.
RH7/SEG43	RH7	0	0	DIG	LATH<7> data output.
		1	I	ST	PORTH<7> data input.
	SEG43	x	0	ANA	LCD Segment 43 output; disables all other pin functions.

TABLE 10-18: PORTH FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
PORTH	RH7	RH6	RH5	RH4	RH3	RH2	RH1	RH0	62
LATH	LATH7	LATH6	LATH5	LATH4	LATH3	LATH2	LATH1	LATH0	62
TRISH	TRISH7	TRISH6	TRISH5	TRISH4	TRISH3	TRISH2	TRISH1	TRISH0	62
LCDSE5	SE47	SE46	SE45	SE44	SE43	SE42	SE41	SE40	61

10.10 PORTJ, TRISJ and LATJ Registers

Note:	PORTJ is available only on PIC18F8XJ90
	devices.

PORTJ is an 8-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISJ and LATJ. All pins on PORTJ are digital only and tolerate voltages up to 5.5V.

All pins on PORTJ are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output.

Note:	These pins are configured as digital inputs
	on any device Reset.

All PORTJ pins, except RJ0, are multiplexed with LCD segment drives controlled by the LCDSE4 register. I/O port functions are only available on these pins when the segments are disabled.

Each of the PORTJ pins has a weak internal pull-up. The pull-ups are provided to keep the inputs at a known state for the external memory interface while powering up. A single control bit can turn off all the pull-ups. This is performed by clearing bit, RJPU (PORTG<5>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on any device Reset.

EXAMPLE 10-9:	INITIALIZING PORTJ
LAAWII LL IV-J.	

CLRF	PORTJ	; Initialize PORTJ by
		; clearing output latches
CLRF	LATJ	; Alternate method
		; to clear output latches
MOVLW	0CFh	; Value used to
		; initialize data
		; direction
MOVWF	TRISJ	; Set RJ3:RJ0 as inputs
		; RJ5:RJ4 as output
		; RJ7:RJ6 as inputs

Pin Name	Function	TRIS Setting	I/O	I/O Type	Description
RJ0	RJ0	0	0	DIG	LATJ<0> data output.
		1	Ι	ST	PORTJ<0> data input.
RJ1/SEG33	RJ1	0	0	DIG	LATJ<1> data output.
		1	Ι	ST	PORTJ<1> data input.
	SEG33	x	0	ANA	LCD Segment 33 output; disables all other pin functions.
RJ2/SEG34	RJ2	0	0	DIG	LATJ<2> data output.
		1		ST	PORTJ<2> data input.
	SEG34	x	0	ANA	LCD Segment 34 output; disables all other pin functions.
RJ3/SEG35 RJ3 SEG35		0	0	DIG	LATJ<3> data output.
		1	-	ST	PORTJ<3> data input.
		x	0	ANA	LCD Segment 35 output; disables all other pin functions.
RJ4/SEG39	RJ4	0	0	DIG	LATJ<4> data output.
		1	-	ST	PORTJ<4> data input.
	SEG39	x	0	ANA	LCD Segment 39 output; disables all other pin functions.
RJ5/SEG38	RJ5	0	0	DIG	LATJ<5> data output.
		1	-	ST	PORTJ<5> data input.
	SEG38	x	0	ANA	LCD Segment 38 output; disables all other pin functions.
RJ6/SEG37	RJ6	0	0	DIG	LATJ<6> data output.
		1	Ι	ST	PORTJ<6> data input.
	SEG37	x	0	ANA	LCD Segment 37 output; disables all other pin functions.
RJ7/SEG36	RJ7	0	0	DIG	LATJ<7> data output.
		1	Ι	ST	PORTJ<7> data input.
	SEG36	x	0	ANA	LCD Segment 36 output; disables all other pin functions.

TABLE 10-20: PORTJ FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

TABLE 10-21: SUMMARY OF REGISTERS ASSOCIATED WITH PORTJ

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
PORTJ	RJ7	RJ6	RJ5	RJ4	RJ3	RJ2	RJ1	RJ0	62
LATJ	LATJ7	LATJ6	LATJ5	LATJ4	LATJ3	LATJ2	LATJ1	LATJ0	62
TRISJ	TRISJ7	TRISJ6	TRISJ5	TRISJ4	TRISJ3	TRISJ2	TRISJ1	TRISJ0	62
PORTG	RDPU	REPU	RJPU ⁽¹⁾	RG4	RG3	RG2	RG1	RG0	62
LCDSE4	SE39	SE38	SE37	SE36	SE35	SE34	SE33	SE32	61

Legend: Shaded cells are not used by PORTJ.

Note 1: Unimplemented on PIC18F6XJ90 devices, read as '0'.

11.0 TIMER0 MODULE

The Timer0 module incorporates the following features:

- Software selectable operation as a timer or counter in both 8-bit or 16-bit modes
- · Readable and writable registers
- Dedicated 8-bit, software programmable
 prescaler
- Selectable clock source (internal or external)
- Edge select for external clock
- Interrupt-on-overflow

The T0CON register (Register 11-1) controls all aspects of the module's operation, including the prescale selection; it is both readable and writable.

A simplified block diagram of the Timer0 module in 8-bit mode is shown in Figure 11-1. Figure 11-2 shows a simplified block diagram of the Timer0 module in 16-bit mode.

REGISTER 11-1: T0CON: TIMER0 CONTROL REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
TMR00N	T08BIT	TOCS	T0SE	PSA	T0PS2	T0PS1	T0PS0
bit 7					•		bit 0

Legend:				
R = Reada	ble bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value	at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown
bit 7		l: Timer0 On/Off Control bit les Timer0 s Timer0		
bit 6	1 = Time	Timer0 8-Bit/16-Bit Control bit 10 is configured as an 8-bit tir 10 is configured as a 16-bit tir	ner/counter	
bit 5	1 = Trans	mer0 Clock Source Select bit ition on T0CKI pin al instruction cycle clock (CL		
bit 4	1 = Incre	ner0 Source Edge Select bit ment on high-to-low transitior ment on low-to-high transitior		
bit 3	1 = TIme		t Timer0 clock input bypasses er0 clock input comes from pr	
bit 2-0	111 = 1:2 110 = 1:1 101 = 1:0 100 = 1:2 011 = 1:1 010 = 1:8	0>: Timer0 Prescaler Select b 256 Prescale value 128 Prescale value 128 Prescale value 129 Prescale value 120 Prescale value 121 Prescale value 122 Prescale value 123 Prescale value 126 Prescale value 127 Prescale value 128 Prescale value 129 Prescale value 129 Prescale value 120 Prescale value 120 Prescale value 121 Prescale value 122 Prescale value 123 Prescale value 124 Prescale value 125 Prescale value 126 Prescale value 127 Prescale value 128 Prescale value 129 Prescale value 129 Prescale value 129 Prescale value 129 Prescale value	pits	

11.1 Timer0 Operation

Timer0 can operate as either a timer or a counter. The mode is selected with the T0CS bit (T0CON<5>). In Timer mode (T0CS = 0), the module increments on every clock by default unless a different prescaler value is selected (see **Section 11.3 "Prescaler"**). If the TMR0 register is written to, the increment is inhibited for the following two instruction cycles. The user can work around this by writing an adjusted value to the TMR0 register.

The Counter mode is selected by setting the T0CS bit (= 1). In this mode, Timer0 increments either on every rising or falling edge of pin, RA4/T0CKI. The incrementing edge is determined by the Timer0 Source Edge Select bit, T0SE (T0CON<4>); clearing this bit selects the rising edge. Restrictions on the external clock input are discussed below.

An external clock source can be used to drive Timer0, however, it must meet certain requirements to ensure that the external clock can be synchronized with the internal phase clock (Tosc). There is a delay between synchronization and the onset of incrementing the timer/counter.

11.2 Timer0 Reads and Writes in 16-Bit Mode

TMR0H is not the actual high byte of Timer0 in 16-bit mode. It is actually a buffered version of the real high byte of Timer0, which is not directly readable nor writable (refer to Figure 11-2). TMR0H is updated with the contents of the high byte of Timer0 during a read of TMR0L. This provides the ability to read all 16 bits of Timer0 without having to verify that the read of the high and low byte were valid, due to a rollover between successive reads of the high and low byte.

Similarly, a write to the high byte of Timer0 must also take place through the TMR0H Buffer register. The high byte is updated with the contents of TMR0H when a write occurs to TMR0L. This allows all 16 bits of Timer0 to be updated at once.

FIGURE 11-1: TIMER0 BLOCK DIAGRAM (8-BIT MODE)

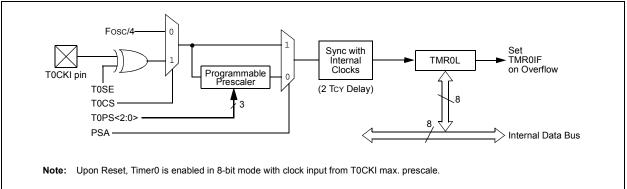
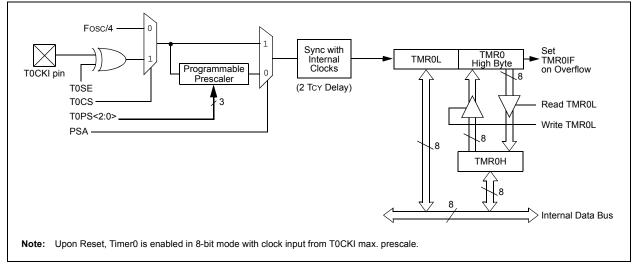


FIGURE 11-2: TIMER0 BLOCK DIAGRAM (16-BIT MODE)



11.3 Prescaler

An 8-bit counter is available as a prescaler for the Timer0 module. The prescaler is not directly readable or writable. Its value is set by the PSA and T0PS<2:0> bits (T0CON<3:0>) which determine the prescaler assignment and prescale ratio.

Clearing the PSA bit assigns the prescaler to the Timer0 module. When it is assigned, prescale values from 1:2 through 1:256, in power-of-2 increments, are selectable.

When assigned to the Timer0 module, all instructions writing to the TMR0 register (e.g., CLRF TMR0, MOVWF TMR0, BSF TMR0, etc.) clear the prescaler count.

Note:	Writing to TMR0 when the prescaler is							
	assigned to Timer0 will clear the prescaler							
	count but will not change the prescaler							
	assignment.							

11.3.1 SWITCHING PRESCALER ASSIGNMENT

The prescaler assignment is fully under software control and can be changed "on-the-fly" during program execution.

11.4 Timer0 Interrupt

The TMR0 interrupt is generated when the TMR0 register overflows from FFh to 00h in 8-bit mode, or from FFFFh to 0000h in 16-bit mode. This overflow sets the TMR0IF flag bit. The interrupt can be masked by clearing the TMR0IE bit (INTCON<5>). Before re-enabling the interrupt, the TMR0IF bit must be cleared in software by the Interrupt Service Routine.

Since Timer0 is shut down in Sleep mode, the TMR0 interrupt cannot awaken the processor from Sleep.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values
TMR0L	Timer0 Register Low Byte								on page 60
TMR0H	Timer0 Register High Byte								60
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59
T0CON	TMR0ON	T08BIT	TOCS	T0SE	PSA	T0PS2	T0PS1	T0PS0	60
TRISA	TRISA7 ⁽¹⁾	TRISA6 ⁽¹⁾	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	62

 TABLE 11-1:
 REGISTERS ASSOCIATED WITH TIMER0

Legend: — = unimplemented, read as '0'. Shaded cells are not used by Timer0.

Note 1: RA<7:6> and their associated latch and direction bits are configured as port pins only when the internal oscillator is selected as the default clock source (FOSC2 Configuration bit = 0); otherwise, they are disabled and these bits read as '0'.

NOTES:

12.0 TIMER1 MODULE

The Timer1 timer/counter module incorporates these features:

- Software selectable operation as a 16-bit timer or counter
- Readable and writable 8-bit registers (TMR1H and TMR1L)
- Selectable clock source (internal or external) with device clock or Timer1 oscillator internal options
- Interrupt-on-overflow
- Reset on CCP Special Event Trigger
- Device clock status flag (T1RUN)

A simplified block diagram of the Timer1 module is shown in Figure 12-1. A block diagram of the module's operation in Read/Write mode is shown in Figure 12-2.

The module incorporates its own low-power oscillator to provide an additional clocking option. The Timer1 oscillator can also be used as a low-power clock source for the microcontroller in power-managed operation.

Timer1 can also be used to provide Real-Time Clock (RTC) functionality to applications with only a minimal addition of external components and code overhead.

Timer1 is controlled through the T1CON Control register (Register 12-1). It also contains the Timer1 Oscillator Enable bit (T1OSCEN). Timer1 can be enabled or disabled by setting or clearing control bit, TMR1ON (T1CON<0>).

R/W-0	R-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
RD16	T1RUN	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N
bit 7							bit 0

REGISTER 12-1: T1CON: TIMER1 CONTROL REGISTER

Legend:							
R = Readable bit		W = Writable bit	U = Unimplemented bit, read as '0'				
-n = Value	at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknowr			
bit 7	RD16: 1	6-Bit Read/Write Mode Enab	le bit				
		bles register read/write of Tli bles register read/write of Tir	mer1 in one 16-bit operation mer1 in two 8-bit operations				
bit 6	T1RUN:	Timer1 System Clock Status	s bit				
	-	ice clock is derived from Tim ice clock is derived from ano					
bit 5-4	T1CKPS<1:0>: Timer1 Input Clock Prescale Select bits						
	10 = 1:4 01 = 1:2	Prescale value Prescale value Prescale value Prescale value					
bit 3	T1OSCE	N: Timer1 Oscillator Enable	bit				
		r1 oscillator is enabled					
		r1 oscillator is shut off	resistor are turned off to elimina	to power drain			
bit 2			at Synchronization Select bit				
		IR1CS = 1:	at Synchronization Select bit				
	1 = Do n	ot synchronize external clock hronize external clock	k input				
		<u>/IR1CS = 0:</u> s ignored. Timer1 uses the ir	nternal clock when TMR1CS =	0.			
bit 1	TMR1CS	: Timer1 Clock Source Sele	ct bit				
		rnal clock from pin RC0/T1C nal clock (Fosc/4)	DSO/T13CKI (on the rising edge	e)			
bit 0	TMR10	I: Timer1 On bit					
	1 = Ena 0 = Stop	bles Timer1					

12.1 Timer1 Operation

Timer1 can operate in one of these modes:

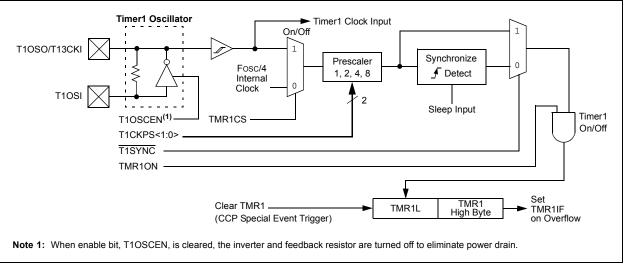
- Timer
- Synchronous Counter
- Asynchronous Counter

The operating mode is determined by the clock select bit, TMR1CS (T1CON<1>). When TMR1CS is cleared (= 0), Timer1 increments on every internal instruction

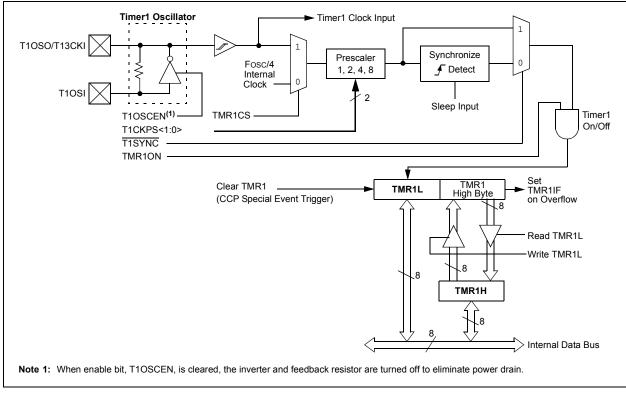
cycle (Fosc/4). When the bit is set, Timer1 increments on every rising edge of the Timer1 external clock input or the Timer1 oscillator, if enabled.

When Timer1 is enabled, the RC1/T1OSI and RC0/T1OSO/T13CKI pins become inputs. This means the values of TRISC<1:0> are ignored and the pins are read as '0'.

FIGURE 12-1: TIMER1 BLOCK DIAGRAM (8-BIT MODE)







12.2 Timer1 16-Bit Read/Write Mode

Timer1 can be configured for 16-bit reads and writes (see Figure 12-2). When the RD16 control bit (T1CON<7>) is set, the address for TMR1H is mapped to a buffer register for the high byte of Timer1. A read from TMR1L will load the contents of the high byte of Timer1 into the Timer1 High Byte Buffer register. This provides the user with the ability to accurately read all 16 bits of Timer1 without having to determine whether a read of the high byte, followed by a read of the low byte, has become invalid due to a rollover between reads.

A write to the high byte of Timer1 must also take place through the TMR1H Buffer register. The Timer1 high byte is updated with the contents of TMR1H when a write occurs to TMR1L. This allows a user to write all 16 bits to both the high and low bytes of Timer1 at once.

The high byte of Timer1 is not directly readable or writable in this mode. All reads and writes must take place through the Timer1 High Byte Buffer register. Writes to TMR1H do not clear the Timer1 prescaler. The prescaler is only cleared on writes to TMR1L.

12.3 Timer1 Oscillator

An on-chip crystal oscillator circuit is incorporated between pins, T1OSI (input) and T1OSO (amplifier output). It is enabled by setting the Timer1 Oscillator Enable bit, T1OSCEN (T1CON<3>). The oscillator is a low-power circuit rated for 32 kHz crystals. It will continue to run during all power-managed modes. The circuit for a typical LP oscillator is shown in Figure 12-3. Table 12-1 shows the capacitor selection for the Timer1 oscillator.

The user must provide a software time delay to ensure proper start-up of the Timer1 oscillator.

FIGURE 12-3: EXTERNAL COMPONENTS FOR THE TIMER1 LP OSCILLATOR

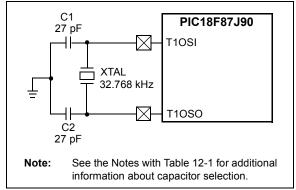


TABLE 12-1: CAPACITOR SELECTION FOR THE TIMER1 OSCILLATOR^(2,3,4)

	USCILLATOR								
Oscillator Type	Freq.	C1	C2						
LP	32.768 kHz	27 pF ⁽¹⁾	27 pF ⁽¹⁾						
Note 1: Microchip suggests these values as a starting point in validating the oscillator circuit.									
(Higher capacita of the oscillate start-up time.		-						
t t	Since each resonator/crystal has its own characteristics, the user should consult the resonator/crystal manufacturer for appropriate values of external components.								
	Capacitor valu	es are for des	ign guidance						

12.3.1 USING TIMER1 AS A CLOCK SOURCE

The Timer1 oscillator is also available as a clock source in power-managed modes. By setting the System Clock Select bits, SCS<1:0> (OSCCON<1:0>), to '01', the device switches to SEC_RUN mode; both the CPU and peripherals are clocked from the Timer1 oscillator. If the IDLEN bit (OSCCON<7>) is cleared and a SLEEP instruction is executed, the device enters SEC_IDLE mode. Additional details are available in **Section 4.0 "Power-Managed Modes"**.

Whenever the Timer1 oscillator is providing the clock source, the Timer1 System Clock Status Flag, T1RUN (T1CON<6>), is set. This can be used to determine the controller's current clocking mode. It can also indicate the clock source being currently used by the Fail-Safe Clock Monitor. If the Clock Monitor is enabled and the Timer1 oscillator fails while providing the clock, polling the T1RUN bit will indicate whether the clock is being provided by the Timer1 oscillator or another source.

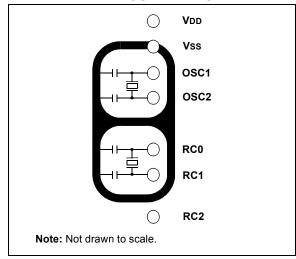
12.3.2 TIMER1 OSCILLATOR LAYOUT CONSIDERATIONS

The Timer1 oscillator circuit draws very little power during operation. Due to the low-power nature of the oscillator, it may also be sensitive to rapidly changing signals in close proximity.

The oscillator circuit, shown in Figure 12-3, should be located as close as possible to the microcontroller. There should be no circuits passing within the oscillator circuit boundaries other than Vss or VDD.

If a high-speed circuit must be located near the oscillator (such as the CCP1 pin in Output Compare or PWM mode, or the primary oscillator using the OSC2 pin), a grounded guard ring around the oscillator circuit, as shown in Figure 12-4, may be helpful when used on a single-sided PCB or in addition to a ground plane.

FIGURE 12-4: OSCILLATOR CIRCUIT WITH GROUNDED GUARD RING



12.4 Timer1 Interrupt

The TMR1 register pair (TMR1H:TMR1L) increments from 0000h to FFFFh and rolls over to 0000h. The Timer1 interrupt, if enabled, is generated on overflow which is latched in interrupt flag bit, TMR1IF (PIR1<0>). This interrupt can be enabled or disabled by setting or clearing the Timer1 Interrupt Enable bit, TMR1IE (PIE1<0>).

12.5 Resetting Timer1 Using the CCP Special Event Trigger

If CCP1 or CCP2 is configured to use Timer1 and to generate a Special Event Trigger in Compare mode (CCPxM<3:0> = 1011), this signal will reset Timer3. The trigger from CCP2 will also start an A/D conversion if the A/D module is enabled (see Section 16.3.4 "Special Event Trigger" for more information).

The module must be configured as either a timer or a synchronous counter to take advantage of this feature. When used this way, the CCPRxH:CCPRxL register pair effectively becomes a period register for Timer1.

If Timer1 is running in Asynchronous Counter mode, this Reset operation may not work.

In the event that a write to Timer1 coincides with a Special Event Trigger, the write operation will take precedence.

Note:	The Special Event Triggers from the CCPx
	module will not set the TMR1IF interrupt
	flag bit (PIR1<0>).

12.6 Using Timer1 as a Real-Time Clock

Adding an external LP oscillator to Timer1 (such as the one described in **Section 12.3 "Timer1 Oscillator**") gives users the option to include RTC functionality to their applications. This is accomplished with an inexpensive watch crystal to provide an accurate time base and several lines of application code to calculate the time. When operating in Sleep mode and using a battery or supercapacitor as a power source, it can completely eliminate the need for a separate RTC device and battery backup.

The application code routine, RTCisr, shown in Example 12-1, demonstrates a simple method to increment a counter, at one-second intervals, using an Interrupt Service Routine. Incrementing the TMR1 register pair to overflow, triggers the interrupt and calls the routine which increments the seconds counter by one. Additional counters for minutes and hours are incremented as the previous counter overflows.

Since the register pair is 16 bits wide, counting up to overflow the register directly from a 32.768 kHz clock would take 2 seconds. To force the overflow at the required one-second intervals, it is necessary to preload it. The simplest method is to set the MSb of TMR1H with a BSF instruction. Note that the TMR1L register is never preloaded or altered; doing so may introduce cumulative error over many cycles.

For this method to be accurate, Timer1 must operate in Asynchronous mode and the Timer1 overflow interrupt must be enabled (PIE1<0> = 1) as shown in the routine, RTCinit. The Timer1 oscillator must also be enabled and running at all times.

RTCinit			
	MOVLW	80h	; Preload TMR1 register pair
	MOVWF	TMR1H	; for 1 second overflow
	CLRF	TMR1L	
	MOVLW	b'00001111'	; Configure for external clock,
	MOVWF	T1CON	; Asynchronous operation, external oscillator
	CLRF	secs	; Initialize timekeeping registers
	CLRF	mins	;
	MOVLW	.12	
	MOVWF	hours	
	BSF	PIE1, TMR1IE	; Enable Timer1 interrupt
	RETURN		
RTCisr			
	BSF	TMR1H, 7	; Preload for 1 sec overflow
	BCF	PIR1, TMR1IF	; Clear interrupt flag
	INCF	secs, F	; Increment seconds
	MOVLW	.59	; 60 seconds elapsed?
	CPFSGT	secs	
	RETURN		; No, done
	CLRF	secs	; Clear seconds
	INCF	mins, F	; Increment minutes
	MOVLW	.59	; 60 minutes elapsed?
	CPFSGT	mins	
	RETURN		; No, done
	CLRF	mins	; clear minutes
	INCF	hours, F	; Increment hours
	MOVLW	.23	; 24 hours elapsed?
	CPFSGT	hours	
	RETURN		; No, done
	CLRF	hours	; Reset hours
	RETURN		; Done

TABLE 12-2: REGISTERS ASSOCIATED WITH TIMER1 AS A TIMER/COUNTER

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59
PIR1	—	ADIF	RC1IF	TX1IF	SSPIF	_	TMR2IF	TMR1IF	62
PIE1	—	ADIE	RC1IE	TX1IE	SSPIE		TMR2IE	TMR1IE	62
IPR1	_	ADIP	RC1IP	TX1IP	SSPIP		TMR2IP	TMR1IP	62
TMR1L	Timer1 Reg	gister Low By	/te						60
TMR1H	H Timer1 Register High Byte								60
T1CON	RD16	T1RUN	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N	60

Legend: Shaded cells are not used by the Timer1 module.

NOTES:

13.0 TIMER2 MODULE

The Timer2 module incorporates the following features:

- 8-bit Timer and Period registers (TMR2 and PR2, respectively)
- Readable and writable (both registers)
- Software programmable prescaler (1:1, 1:4 and 1:16)
- Software programmable postscaler (1:1 through 1:16)
- Interrupt on TMR2 to PR2 match
- Optional use as the shift clock for the MSSP module

The module is controlled through the T2CON register (Register 13-1) which enables or disables the timer, and configures the prescaler and postscaler. Timer2 can be shut off by clearing control bit, TMR2ON (T2CON<2>), to minimize power consumption.

A simplified block diagram of the module is shown in Figure 13-1.

13.1 Timer2 Operation

In normal operation, TMR2 is incremented from 00h on each clock (Fosc/4). A 4-bit counter/prescaler on the clock input gives direct input, divide-by-4 and divide-by-16 prescale options. These are selected by the prescaler control bits, T2CKPS<1:0> (T2CON<1:0>). The value of TMR2 is compared to that of the Period register, PR2, on each clock cycle. When the two values match, the comparator generates a match signal as the timer output. This signal also resets the value of TMR2 to 00h on the next cycle and drives the output counter/postscaler (see **Section 13.2 "Timer2 Interrupt"**).

The TMR2 and PR2 registers are both directly readable and writable. The TMR2 register is cleared on any device Reset, while the PR2 register initializes at FFh. Both the prescaler and postscaler counters are cleared on the following events:

- a write to the TMR2 register
- a write to the T2CON register
- any device Reset (Power-on Reset, MCLR Reset, Watchdog Timer Reset or Brown-out Reset)

TMR2 is not cleared when T2CON is written.

REGISTER 13-1: T2CON: TIMER2 CONTROL REGISTER

U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	T2OUTPS3	T2OUTPS2	T2OUTPS1	T2OUTPS0	TMR2ON	T2CKPS1	T2CKPS0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	Unimplemented: Read as '0'
bit 6-3	T2OUTPS<3:0>: Timer2 Output Postscale Select bits
	0000 = 1:1 Postscale
	0001 = 1:2 Postscale
	•
	•
	•
	1111 = 1:16 Postscale
bit 2	TMR2ON: Timer2 On bit
	1 = Timer2 is on
	0 = Timer2 is off
bit 1-0	T2CKPS<1:0>: Timer2 Clock Prescale Select bits
	00 = Prescaler is 1
	01 = Prescaler is 4
	1x = Prescaler is 16

13.2 Timer2 Interrupt

Timer2 can also generate an optional device interrupt. The Timer2 output signal (TMR2 to PR2 match) provides the input for the 4-bit output counter/postscaler. This counter generates the TMR2 match interrupt flag which is latched in TMR2IF (PIR1<1>). The interrupt is enabled by setting the TMR2 Match Interrupt Enable bit, TMR2IE (PIE1<1>).

A range of 16 postscale options (from 1:1 through 1:16 inclusive) can be selected with the postscaler control bits, T2OUTPS<3:0> (T2CON<6:3>).

13.3 Timer2 Output

The unscaled output of TMR2 is available primarily to the CCP modules, where it is used as a time base for operations in PWM mode.

Timer2 can be optionally used as the shift clock source for the MSSP module operating in SPI mode. Additional information is provided in Section 18.0 "Master Synchronous Serial Port (MSSP) Module".

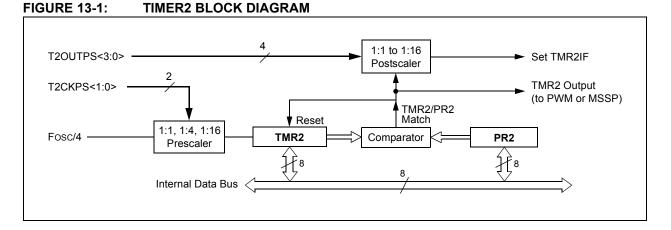


TABLE 13-1: REGISTERS ASSOCIATED WITH TIMER2 AS A TIMER/COUNTER

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page	
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59	
PIR1	_	ADIF	RC1IF	TX1IF	SSPIF	_	TMR2IF	TMR1IF	62	
PIE1	—	ADIE	RC1IE	TX1IE	SSPIE	_	TMR2IE	TMR1IE	62	
IPR1	—	ADIP	RC1IP	TX1IP	SSPIP	—	TMR2IP	TMR1IP	62	
TMR2	Timer2 Reg	gister							60	
T2CON	_	- T2OUTPS3 T2OUTPS2 T2OUTPS1 T2OUTPS0 TMR2ON T2CKPS1 T2CKPS0								
PR2	Timer2 Per	iod Register							60	

Legend: — = unimplemented, read as '0'. Shaded cells are not used by the Timer2 module.

14.0 TIMER3 MODULE

The Timer3 timer/counter module incorporates these features:

- Software selectable operation as a 16-bit timer or counter
- Readable and writable 8-bit registers (TMR3H and TMR3L)
- Selectable clock source (internal or external) with device clock or Timer1 oscillator internal options
- Interrupt on overflow
- Module Reset on CCP Special Event Trigger

REGISTER 14-1: T3CON: TIMER3 CONTROL REGISTER

A simplified block diagram of the Timer3 module is shown in Figure 14-1. A block diagram of the module's operation in Read/Write mode is shown in Figure 14-2.

The Timer3 module is controlled through the T3CON register (Register 14-1). It also selects the clock source options for the CCP modules. See **Section 16.2.2** "**Timer1/Timer3 Mode Selection**" for more information.

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
RD16	T3CCP2	T3CKPS1	T3CKPS0	T3CCP1	T3SYNC	TMR3CS	TMR3ON
bit 7							bit 0

Legend:			
R = Readab	le bit W = Writable I	bit U = Unimplemented bi	it, read as '0'
-n = Value a	t POR '1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown
bit 7	RD16: 16-Bit Read/Write Mo	de Enable bit	
		te of Timer3 in one 16-bit operation te of Timer3 in two 8-bit operations	
bit 6,3	T3CCP<2:1>: Timer3 and Ti	ner1 to CCPx Enable bits	
	01 = Timer3 is the capture/co Timer1 is the capture/co	ompare clock source for the CCP mod ompare clock source for CCP2; ompare clock source for CCP1 ompare clock source for the CCP mod	
bit 5-4	T3CKPS<1:0>: Timer3 Input	Clock Prescale Select bits	
	11 = 1:8 Prescale value 10 = 1:4 Prescale value 01 = 1:2 Prescale value 00 = 1:1 Prescale value		
bit 2		ock Input Synchronization Control bit k comes from Timer1/Timer3.)	
	When TMR3CS = 1: 1 = Does not synchronize the 0 = Synchronizes the externa		
	<u>When TMR3CS = 0:</u> This bit is ignored. Timer3 us	es the internal clock when TMR3CS =	= 0.
bit 1	TMR3CS: Timer3 Clock Sou	rce Select bit	
	falling edge)	n Timer1 oscillator or T13CKI (on the r	ising edge after the first
	0 = Internal clock (Fosc/4)		
bit 0	TMR3ON: Timer3 On bit 1 = Enables Timer3 0 = Stops Timer3		

14.1 Timer3 Operation

Timer3 can operate in one of three modes:

- Timer
- Synchronous Counter
- Asynchronous Counter

The operating mode is determined by the clock select bit, TMR3CS (T3CON<1>). When TMR3CS is cleared (= 0), Timer3 increments on every internal instruction cycle (Fosc/4). When the bit is set, Timer3 increments on every rising edge of the Timer1 external clock input or the Timer1 oscillator, if enabled.

As with Timer1, the RC1/T1OSI and RC0/T1OSO/T13CKI pins become inputs when the Timer1 oscillator is enabled. This means the values of TRISC<1:0> are ignored and the pins are read as '0'.

FIGURE 14-1: TIMER3 BLOCK DIAGRAM (8-BIT MODE)

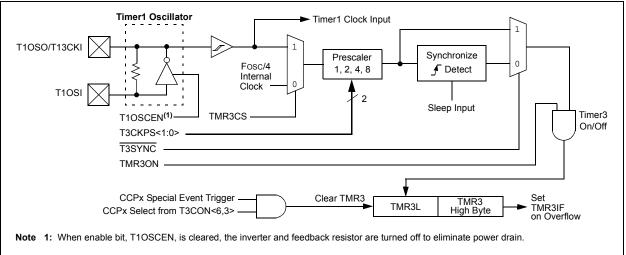
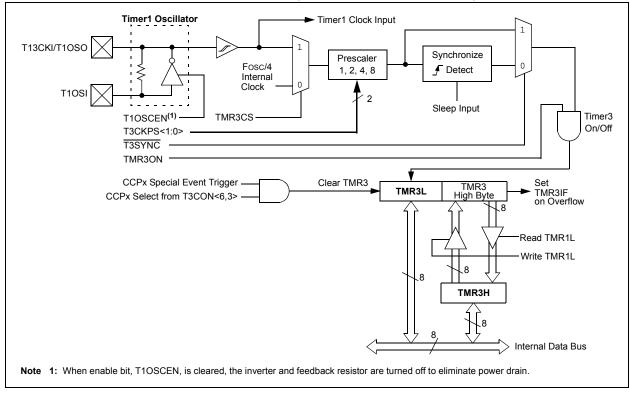


FIGURE 14-2: TIMER3 BLOCK DIAGRAM (16-BIT READ/WRITE MODE)



14.2 Timer3 16-Bit Read/Write Mode

Timer3 can be configured for 16-bit reads and writes (see Figure 14-2). When the RD16 control bit (T3CON<7>) is set, the address for TMR3H is mapped to a buffer register for the high byte of Timer3. A read from TMR3L will load the contents of the high byte of Timer3 into the Timer3 High Byte Buffer register. This provides the user with the ability to accurately read all 16 bits of Timer1 without having to determine whether a read of the high byte, followed by a read of the low byte, has become invalid due to a rollover between reads.

A write to the high byte of Timer3 must also take place through the TMR3H Buffer register. The Timer3 high byte is updated with the contents of TMR3H when a write occurs to TMR3L. This allows a user to write all 16 bits to both the high and low bytes of Timer3 at once.

The high byte of Timer3 is not directly readable or writable in this mode. All reads and writes must take place through the Timer3 High Byte Buffer register.

Writes to TMR3H do not clear the Timer3 prescaler. The prescaler is only cleared on writes to TMR3L.

14.3 Using the Timer1 Oscillator as the Timer3 Clock Source

The Timer1 internal oscillator may be used as the clock source for Timer3. The Timer1 oscillator is enabled by setting the T1OSCEN (T1CON<3>) bit. To use it as the Timer3 clock source, the TMR3CS bit must also be set. As previously noted, this also configures Timer3 to increment on every rising edge of the oscillator source.

The Timer1 oscillator is described in Section 12.0 "Timer1 Module".

14.4 Timer3 Interrupt

The TMR3 register pair (TMR3H:TMR3L) increments from 0000h to FFFFh and overflows to 0000h. The Timer3 interrupt, if enabled, is generated on overflow and is latched in interrupt flag bit, TMR3IF (PIR2<1>). This interrupt can be enabled or disabled by setting or clearing the Timer3 Interrupt Enable bit, TMR3IE (PIE2<1>).

14.5 Resetting Timer3 Using the CCP Special Event Trigger

If CCP1 or CCP2 is configured to use Timer3 and to generate a Special Event Trigger in Compare mode (CCPxM<3:0> = 1011), this signal will reset Timer3. The trigger from CCP2 will also start an A/D conversion if the A/D module is enabled (see **Section 16.3.4** "**Special Event Trigger**" for more information).

The module must be configured as either a timer or synchronous counter to take advantage of this feature. When used this way, the CCPRxH:CCPRxL register pair effectively becomes a period register for Timer3.

If Timer3 is running in Asynchronous Counter mode, the Reset operation may not work.

In the event that a write to Timer3 coincides with a Special Event Trigger from a CCP module, the write will take precedence.

Note: The Special Event Triggers from the CCPx module will not set the TMR3IF interrupt flag bit (PIR2<1>).

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59
PIR2	OSCFIF	CMIF	_	_	BCLIF	LVDIF	TMR3IF	_	62
PIE2	OSCFIE	CMIE	_	_	BCLIE	LVDIE	TMR3IE	_	62
IPR2	OSCFIP	CMIP	_	_	BCLIP	LVDIP	TMR3IP	—	62
TMR3L	Timer3 Reg	gister Low B	yte						61
TMR3H	Timer3 Reg	gister High B	yte						61
T1CON	RD16	T1RUN	T1CKPS1	T1CKPS0	T10SCEN	T1SYNC	TMR1CS	TMR10N	60
T3CON	RD16	T3CCP2	T3CKPS1	T3CKPS0	T3CCP1	T3SYNC	TMR3CS	TMR3ON	61
Lawawala			1 (1) 01				0 1 1		

TABLE 14-1: REGISTERS ASSOCIATED WITH TIMER3 AS A TIMER/COUNTER

Legend: — = unimplemented, read as '0'. Shaded cells are not used by the Timer3 module.

NOTES:

15.0 REAL-TIME CLOCK AND CALENDAR (RTCC)

The key features of the Real-Time Clock and Calendar (RTCC) module are:

- Time: hours, minutes and seconds
- 24-hour format (military time)
- Calendar: weekday, date, month and year
- Alarm configurable
- Year range: 2000 to 2099
- · Leap year correction
- BCD format for compact firmware
- · Optimized for low-power operation
- · User calibration with auto-adjust
- Calibration range: ±2.64 seconds error per month
- Requirements: external 32.768 kHz clock crystal
- Alarm pulse or seconds clock output on RTCC pin

The RTCC module is intended for applications where accurate time must be maintained for an extended period, with minimum to no intervention from the CPU. The module is optimized for low-power usage in order to provide extended battery life while keeping track of time.

The module is a 100-year clock and calendar with automatic leap year detection. The range of the clock is from 00:00:00 (midnight) on January 1, 2000 to 23:59:59 on December 31, 2099. Hours are measured in 24-hour (military time) format. The clock provides a granularity of one second with half-second visibility to the user.

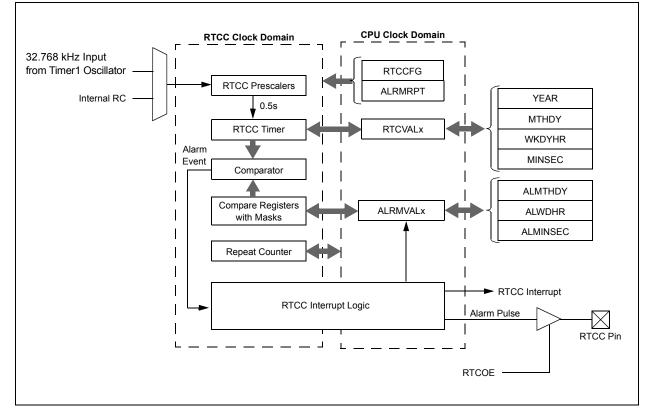


FIGURE 15-1: RTCC BLOCK DIAGRAM

15.1 RTCC MODULE REGISTERS

The RTCC module registers are divided into following categories:

RTCC Control Registers

- RTCCFG
- RTCCAL
- PADCFG1
- ALRMCFG
- ALRMRPT

RTCC Value Registers

- RTCVALH and RTCVALL Can access the following registers
 - YEAR
 - MONTH
 - DAY
 - WEEKDAY
 - HOUR
 - MINUTE
 - SECOND

Alarm Value Registers

- ALRMVALH and ALRMVALL Can access the following registers:
 - ALRMMNTH
 - ALRMDAY
 - ALRMWD
 - ALRMHR
 - ALRMMIN
 - ALRMSEC
- Note: The RTCVALH and RTCVALL registers can be accessed through RTCRPT<1:0>. ALRMVALH and ALRMVALL can be accessed through ALRMPTR<1:0>.

15.1.1 RTCC CONTROL REGISTERS

REGISTER 15-1: RTCCFG: RTCC CONFIGURATION REGISTER⁽¹⁾

R/W-0	U-0	R/W-0	R-0	R-0	R/W-0	R/W-0	R/W-0
RTCEN ⁽²⁾		RTCWREN	RTCSYNC	HALFSEC ⁽³⁾	RTCOE	RTCPTR1	RTCPTR0
oit 7							bit (
Legend:	hit	\// - \//ritabla	hit.	II – Unimplom	opted bit rea	d oo 'O'	
R = Readable		W = Writable '1' = Bit is set		U = Unimplem			
n = Value at	PUR	I = BILIS SEL		'0' = Bit is clea	irea	x = Bit is unkr	IOWN
bit 7	RTCEN: RTC	C Enable bit ⁽²⁾)				
		odule is enable					
	0 = RTCC m	odule is disable	ed				
bit 6	Unimplemen	ted: Read as '	0'				
bit 5	RTCWREN:	RTCC Value Re	egisters Write	Enable bit			
			U U	an be written to l			
	0 = RTCVAL	H and RTCVAL	L registers ar	e locked out fro	m being writte	en to by the use	r
oit 4			•	Synchronizatior			
				registers can c			
		ssumed to be v		register is read	twice and res	uits in the same	data, the dat
				PT registers ca	n be read wit	thout concern o	over a rollove
	ripple	,		0			
bit 3	HALFSEC: ⊦	lalf-Second Sta	tus bit ⁽³⁾				
		half period of a					
		period of a sec					
bit 2		CC Output Enat					
		ock output enal					
		ock output disa					
bit 1-0			-	ndow Pointer bit			/ALL registers
				gisters when rea every read or wri			
	RTCVALH:						
	00 = Minutes						
	01 = Weekda	ıy					
	10 = Month						
	11 = Reserve	ea					
	RTCVALL: 00 = Second:	e					
	01 = Hours	<u> </u>					
	10 = Day						
	11 = Year						
	e RTCCFG regi en if the device		cted by a PO	R. For resets ot	her than POR	, RTCC will con	tinue to run

- 2: A write to the RTCEN bit is only allowed when RTCWREN = 1.
- **3:** This bit is read-only; it is cleared to '0' on a write to the lower half of the MINSEC register.

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
CAL7	CAL6	CAL5	CAL4	CAL3	CAL2	CAL1	CAL0
bit 7							bit 0
Legend:							
R = Readabl	e bit	W = Writable	bit	U = Unimplei	mented bit, read	d as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown
	•			m, auto 500 K			
	01111111 =	Maximum posit	ive adjustme	nt; adds 508 R	TC clock pulses	s every minute	
		Minimum positi No adjustment	•	nt; adds four R ⁻	FC clock pulses	s every minute	
	11111111 =	Minimum negat	tive adjustme	ent; subtracts fo	our RTC clock p	ulses every mir	nute
	•						

REGISTER 15-2: RTCCAL: RTCC CALIBRATION REGISTER

10000000 = Maximum negative adjustment; subtracts 512 RTC clock pulses every minute

REGISTER 15-3: PADCFG1: PAD CONFIGURATION REGISTER

U-0	U-0	U-0	U-0	U-0	R/W-0	R/W-0	U-0
—	—	—	—	_	RTSECSEL1 ⁽¹⁾	RTSECSEL0 ⁽¹⁾	—
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-3 Unimplemented: Read as '0'

bit 2-1 RTSECSEL<1:0>: RTCC Seconds Clock Output Select bit⁽¹⁾

- 11 = Reserved; do not use
 - 10 = RTCC source clock is selected for the RTCC pin (pin can be INTOSC or Timer1 oscillator, depending on the RTCOSC (CONFIG3L<1>) bit setting)⁽²⁾
 - 01 = RTCC seconds clock is selected for the RTCC pin
 - 00 = RTCC alarm pulse is selected for the RTCC pin
- bit 0 Unimplemented: Read as '0'
- Note 1: To enable the actual RTCC output, the RTCOE (RTCCFG<2>) bit must be set.
 - 2: If the Timer1 oscillator is the clock source for RTCC, the T1OSCEN bit should be set (T1CON<3> = 1).

REGISTER 15-4: ALRMCFG: ALARM CONFIGURATION REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
ALRMEN	CHIME	AMASK3	AMASK2	AMASK1	AMASK0	ALRMPTR1	ALRMPTR0
bit 7	1	1	l			•	bit 0
Legend:							
R = Readable	e bit	W = Writable	bit	U = Unimpler	mented bit, read	d as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unki	nown
bit 7		arm Enable bit					
	1 = Alarm is and CHI	enabled (cleared) MF = 0)	ed automatica	lly after an alai	m event when	ever ARP1<7:0	> = 00
	0 = Alarm is	,					
bit 6	CHIME: Chin	ne Enable bit					
		enabled; ALRI				n 00h to FFh	
h # 5 0		disabled; ALR		-	ney reach oon		
bit 5-2		>: Alarm Mask	Configuration	DItS			
	0000 = Ever	ry half second					
		y 10 seconds					
	0011 = Ever						
		ry 10 minutes					
	0101 = Ever 0110 = Onc						
	0111 = Onc	•					
	1000 = Onc						
	1001 = Onc	e a year (excep	t when config	ured for Febru	ary 29 th , once e	every four years	6)
		erved – do not					
		erved – do not					
bit 1-0		I:0>: Alarm Val	-				
		<pre>corresponding ALRMPTR<1:</pre>					
	'00'.			ements on ever	ry read of white		
	ALRMVALH:						
	00 = ALRMM	lin					
	01 = ALRMW						
	10 = ALRMM						
	11 = Unimple	emented					
	ALRMVALL: 00 = ALRMS	FC					
	00 = ALRMS 01 = ALRMH						
	10 = ALRMD						

REGISTER 15-5: ALRMRPT: ALARM REPEAT REGISTER

| R/W-0 |
|-------|-------|-------|-------|-------|-------|-------|-------|
| ARPT7 | ARPT6 | ARPT5 | ARPT4 | ARPT3 | ARPT2 | ARPT1 | ARPT0 |
| bit 7 | | | | | | | bit 0 |

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 ARPT<7:0>: Alarm Repeat Counter Value bits 11111111 = Alarm will repeat 255 more times .

> 00000000 = Alarm will not repeat The counter decrements on any alarm event. The counter is prevented from rolling over from 00h to FFh unless CHIME = 1.

15.1.2 RTCVALH AND RTCVALL REGISTER MAPPINGS

REGISTER 15-6: RESERVED REGISTER

U-0	U-0	U-0	U-0	U-0	U-0	U-0	U-0
_	—	—	—	—	—	—	—
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 Unimplemented: Read as '0'

REGISTER 15-7: YEAR: YEAR VALUE REGISTER⁽¹⁾

| R/W-x |
|--------|--------|--------|--------|--------|--------|--------|--------|
| YRTEN3 | YRTEN2 | YRTEN1 | YRTEN0 | YRONE3 | YRONE2 | YRONE1 | YRONE0 |
| bit 7 | | | | | | | bit 0 |

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-4 **YRTEN<3:0>:** Binary Coded Decimal Value of Year's Tens Digit bits Contains a value from 0 to 9.

bit 3-0 **YRONE<3:0>:** Binary Coded Decimal Value of Year's Ones Digit bits Contains a value from 0 to 9.

Note 1: A write to the YEAR register is only allowed when RTCWREN = 1.

REGISTER 15-8: MONTH: MONTH VALUE REGISTER⁽¹⁾

U-0	U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	—	MTHTEN0	MTHONE3	MTHONE2	MTHONE1	MTHONE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-5	Unimplemented: Read as '0'
bit 4	MTHTEN0: Binary Coded Decimal Value of Month's Tens Digit bit Contains a value of 0 or 1.
bit 3-0	MTHONE<3:0>: Binary Coded Decimal Value of Month's Ones Digit bits Contains a value from 0 to 9.

Note 1: A write to this register is only allowed when RTCWREN = 1.

REGISTER 15-9: DAY: DAY VALUE REGISTER⁽¹⁾

U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	DAYTEN1	DAYTEN0	DAYONE3	DAYONE2	DAYONE1	DAYONE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	1 as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6	Unimplemented: Read as '0'
bit 5-4	DAYTEN<1:0>: Binary Coded Decimal value of Day's Tens Digit bits Contains a value from 0 to 3.
bit 3-0	DAYONE<3:0>: Binary Coded Decimal Value of Day's Ones Digit bits Contains a value from 0 to 9.

Note 1: A write to this register is only allowed when RTCWREN = 1.

REGISTER 15-10: WEEKDAY: WEEKDAY VALUE REGISTER⁽¹⁾

U-0	U-0	U-0	U-0	U-0	R/W-x	R/W-x	R/W-x
—	_			—	WDAY2	WDAY1	WDAY0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-3 Unimplemented: Read as '0'

bit 2-0 **WDAY<2:0>:** Binary Coded Decimal Value of Weekday Digit bits Contains a value from 0 to 6.

Note 1: A write to this register is only allowed when RTCWREN = 1.

REGISTER 15-11: HOUR: HOUR VALUE REGISTER⁽¹⁾

U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	HRTEN1	HRTEN0	HRONE3	HRONE2	HRONE1	HRONE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6	Unimplemented: Read as '0'
bit 5-4	HRTEN<1:0>: Binary Coded Decimal Value of Hour's Tens Digit bits Contains a value from 0 to 2.
bit 3-0	HRONE<3:0>: Binary Coded Decimal Value of Hour's Ones Digit bits Contains a value from 0 to 9.

Note 1: A write to this register is only allowed when RTCWREN = 1.

REGISTER 15-12: MINUTE: MINUTE VALUE REGISTER

U-0	R/W-x						
—	MINTEN2	MINTEN1	MINTEN0	MINONE3	MINONE2	MINONE1	MINONE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	Unimplemented: Read as '0'
bit 6-4	MINTEN<2:0>: Binary Coded Decimal Value of Minute's Tens Digit bits Contains a value from 0 to 5.
bit 3-0	MINONE<3:0>: Binary Coded Decimal Value of Minute's Ones Digit bits Contains a value from 0 to 9.

REGISTER 15-13: SECOND: SECOND VALUE REGISTER

U-0	R/W-x						
—	SECTEN2	SECTEN1	SECTEN0	SECONE3	SECONE2	SECONE1	SECONE0
bit 7							bit 0

Legend:					
R = Readable bit	W = Writable bit	U = Unimplemented bit,	U = Unimplemented bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown		

bit 7	Unimplemented: Read as '0'
bit 6-4	SECTEN<2:0>: Binary Coded Decimal Value of Second's Tens Digit bits Contains a value from 0 to 5.
bit 3-0	SECONE<3:0>: Binary Coded Decimal Value of Second's Ones Digit bits Contains a value from 0 to 9.

15.1.3 ALRMVALH AND ALRMVALL REGISTER MAPPINGS

REGISTER 15-14: ALRMMNTH: ALARM MONTH VALUE REGISTER⁽¹⁾

U-0	U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	
—		—	MTHTEN0	MTHONE3	MTHONE2	MTHONE1	MTHONE0	
bit 7 bit 0								
Legend:								
R = Readable bit W = Writable b		bit	U = Unimpler	nented bit, read	as '0'			

-n = Value at	POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown
bit 7-5	Unimple	emented: Read as '0'		

DIL 7-5	Onimplemented. Read as 0
bit 4	MTHTEN0: Binary Coded Decimal Value of Month's Tens Digit bit Contains a value of 0 or 1.
bit 3-0	MTHONE<3:0>: Binary Coded Decimal Value of Month's Ones Digit bits Contains a value from 0 to 9.

Note 1: A write to this register is only allowed when RTCWREN = 1.

REGISTER 15-15: ALRMDAY: ALARM DAY VALUE REGISTER⁽¹⁾

U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	DAYTEN1	DAYTEN0	DAYONE3	DAYONE2	DAYONE1	DAYONE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6 Unimplemented: Read as '0'	
------------------------------------	--

bit 5-4DAYTEN<1:0>: Binary Coded Decimal Value of Day's Tens Digit bits
Contains a value from 0 to 3.

bit 3-0 **DAYONE<3:0>:** Binary Coded Decimal Value of Day's Ones Digit bits Contains a value from 0 to 9.

Note 1: A write to this register is only allowed when RTCWREN = 1.

REGISTER 15-16: ALRMWD: ALARM WEEKDAY VALUE REGISTER⁽¹⁾

U-0	U-0	U-0	U-0	U-0	R/W-x	R/W-x	R/W-x
—	—	—		_	WDAY2	WDAY1	WDAY0
bit 7							bit 0

Legend:					
R = Readable bit	W = Writable bit	U = Unimplemented bit	U = Unimplemented bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown		

bit 7-3 Unimplemented: Read as '0'

bit 2-0 **WDAY<2:0>:** Binary Coded Decimal Value of Weekday Digit bits Contains a value from 0 to 6.

Note 1: A write to this register is only allowed when RTCWREN = 1.

REGISTER 15-17: ALRMHR: ALARM HOURS VALUE REGISTER⁽¹⁾

U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	HRTEN1	HRTEN0	HRONE3	HRONE2	HRONE1	HRONE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, rea	ad as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6 Unimplemented: Read as '0'

bit 5-4	HRTEN<1:0>: Binary Coded Decimal Value of Hour's Tens Digit bits
	Contains a value from 0 to 2.
bit 3-0	HRONE<3:0>: Binary Coded Decimal Value of Hour's Ones Digit bits
	Contains a value from 0 to 9.

Note 1: A write to this register is only allowed when RTCWREN = 1.

REGISTER 15-18: ALRMMIN: ALARM MINUTES VALUE REGISTER

U-0	R/W-x						
—	MINTEN2	MINTEN1	MINTEN0	MINONE3	MINONE2	MINONE1	MINONE0
bit 7 bit 0							

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit	7	Unimplemented: Read as '0'
bit	6-4	MINTEN<2:0>: Binary Coded Decimal Value of Minute's Tens Digit bits
		Contains a value from 0 to 5.
bit	3-0	MINONE<3:0>: Binary Coded Decimal Value of Minute's Ones Digit bits

Contains a value from 0 to 9.

REGISTER 15-19: ALRMSEC: ALARM SECONDS VALUE REGISTER

U-0	R/W-x						
—	SECTEN2	SECTEN1	SECTEN0	SECONE3	SECONE2	SECONE1	SECONE0
bit 7 bit 0							

Legend:				
R = Reada	ble bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR		'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown
bit 7	Unimple	mented: Read as '0'		
bit 6-4 SECTEN<2:0>: Binary Coded Decimal Value of Second's Tens Digit bits Contains a value from 0 to 5.				
bit 3-0	SECONE<3:0>: Binary Coded Decimal Value of Second's Ones Digit bits			

Contains a value from 0 to 9.

15.1.4 RTCEN BIT WRITE

An attempt to write to the RTCEN bit while RTCWREN = 0 will be ignored. RTCWREN must be set before a write to RTCEN can take place.

Like the RTCEN bit, the RTCVALH and RTCVALL registers can only be written to when RTCWREN = 1. A write to these registers, while RTCWREN = 0, will be ignored.

FIGURE 15-2: TIMER DIGIT FORMAT

15.2 Operation

15.2.1 REGISTER INTERFACE

The register interface for the RTCC and alarm values is implemented using the Binary Coded Decimal (BCD) format. This simplifies the firmware when using the module as each of the digits is contained within its own 4-bit value (see Figure 15-2 and Figure 15-3).

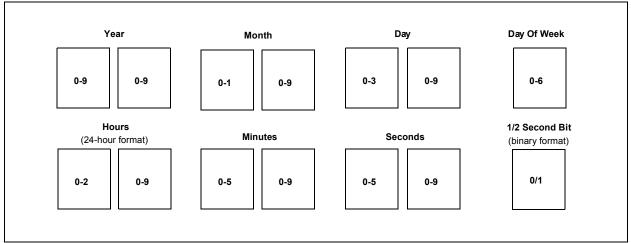
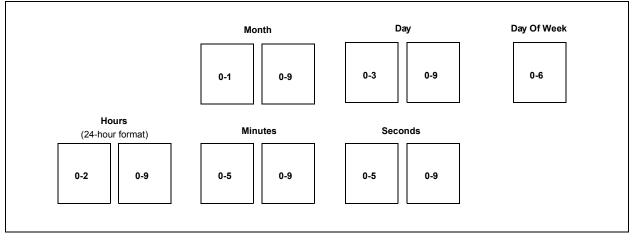


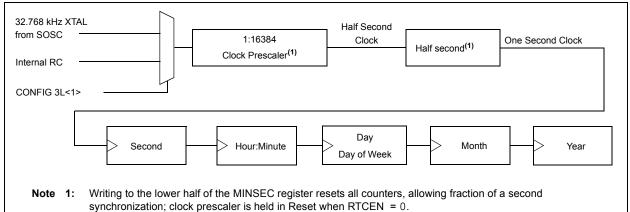
FIGURE 15-3: ALARM DIGIT FORMAT



15.2.2 CLOCK SOURCE

As mentioned earlier, the RTCC module is intended to be clocked by an external Real-Time Clock crystal, oscillating at 32.768 kHz, but can also be an internal oscillator. The RTCC clock selection is decided by the RTCOSC bit (CONFIG3L<1>).

FIGURE 15-4: CLOCK SOURCE MULTIPLEXING



15.2.2.1 Real-Time Clock Enable

The RTCC module can be clocked by an external, 32.768 kHz crystal (Timer1 oscillator) or the Internal RC oscillator, which can be selected in CONFIG3L<1>.

If the external clock is used, the Timer1 oscillator should be enabled by setting the T1OSCEN bit (T1CON<3> = 1). If INTRC is providing the clock, the INTRC clock can be brought out to the RTCC pin by the RTSECSEL<1:0> bits in the PADCFG register.

15.2.3 DIGIT CARRY RULES

This section explains which timer values are affected when there is a rollover.

- Time of Day: from 23:59:59 to 00:00:00 with a carry to the Day field
- Month: from 12/31 to 01/01 with a carry to the Year field
- Day of Week: from 6 to 0 with no carry (see Table 15-1)
- Year Carry: from 99 to 00; this also surpasses the use of the RTCC

For the day to month rollover schedule, see Table 15-2.

Considering that the following values are in BCD format, the carry to the upper BCD digit will occur at a count of 10 and not at 16 (SECONDS, MINUTES, HOURS, WEEKDAY, DAYS and MONTHS).

TABLE 15-1: DAY OF WEEK SCHEDULE

Calibration of the crystal can be done through this

module to yield an error of 3 seconds or less per month.

(For further details, see Section 15.2.9 "Calibration".)

Day of Week			
Sunday	0		
Monday	1		
Tuesday	2		
Wednesday	3		
Thursday	4		
Friday	5		
Saturday	6		

TABLE 15-2: DAY TO MONTH ROLLOVER SCHEDULE

Month	Maximum Day Field
01 (January)	31
02 (February)	28 or 29 ⁽¹⁾
03 (March)	31
04 (April)	30
05 (May)	31
06 (June)	30
07 (July)	31
08 (August)	31
09 (September)	30
10 (October)	31
11 (November)	30
12 (December)	31

Note 1: See Section 15.2.4 "Leap Year".

15.2.4 LEAP YEAR

Since the year range on the RTCC module is 2000 to 2099, the leap year calculation is determined by any year divisible by 4 in the above range. Only February is effected in a leap year.

February will have 29 days in a leap year and 28 days in any other year.

15.2.5 GENERAL FUNCTIONALITY

All Timer registers containing a time value of seconds or greater are writable. The user configures the time by writing the required year, month, day, hour, minutes and seconds to the Timer registers, via register pointers (see **Section 15.2.8 "Register Mapping"**).

The timer uses the newly written values and proceeds with the count from the required starting point.

The RTCC is enabled by setting the RTCEN bit (RTCCFG<7>). If enabled while adjusting these registers, the timer still continues to increment. However, any time the MINSEC register is written to, both of the timer prescalers are reset to '0'. This allows fraction of a second synchronization.

The Timer registers are updated in the same cycle as the write instruction's execution by the CPU. The user must ensure that when RTCEN = 1, the updated registers will not be incremented at the same time. This can be accomplished in several ways:

- By checking the RTCSYNC bit (RTCCFG<4>)
- By checking the preceding digits from which a carry can occur
- By updating the registers immediately following the seconds pulse (or alarm interrupt)

The user has visibility to the half-second field of the counter. This value is read-only and can be reset only by writing to the lower half of the SECONDS register.

15.2.6 SAFETY WINDOW FOR REGISTER READS AND WRITES

The RTCSYNC bit indicates a time window during which the RTCC clock domain registers can be safely read and written without concern about a rollover. When RTCSYNC = 0, the registers can be safely accessed by the CPU.

Whether RTCSYNC = 1 or 0, the user should employ a firmware solution to ensure that the data read did not fall on a rollover boundary, resulting in an invalid or partial read. This firmware solution would consist of reading each register twice and then comparing the two values. If the two values matched, then a rollover did not occur.

15.2.7 WRITE LOCK

In order to perform a write to any of the RTCC Timer registers, the RTCWREN bit (RTCCFG<5>) must be set.

To avoid accidental writes to the RTCC Timer register, it is recommended that the RTCWREN bit (RTCCFG<5>) be kept clear at any time other than while writing to it. For the RTCWREN bit to be set, there is only one instruction cycle time window allowed between the 55h/AA sequence and the setting of RTCWREN. For that reason, it is recommended that users follow the code example in Example 15-1.

EXAMPLE 15-1: SETTING THE RTCWREN BIT

movlw	0x55	
movwf	EECON2	
movlw	0xAA	
movwf	EECON2	
bsf	RTCCFG, RTCWREN	

15.2.8 REGISTER MAPPING

To limit the register interface, the RTCC Timer and Alarm Timer registers are accessed through corresponding register pointers. The RTCC Value register window (RTCVALH and RTCVALL) uses the RTCPTR bits (RTCCFG<1:0>) to select the required Timer register pair.

By reading or writing to the RTCVALH register, the RTCC Pointer value (RTCPTR<1:0>) decrements by '1' until it reaches '00'. Once it reaches '00', the MINUTES and SECONDS value will be accessible through RTCVALH and RTCVALL until the pointer value is manually changed.

TABLE 15-3: RTCVALH AND RTCVALL REGISTER MAPPING

RTCPTR<1:0>	RTCC Value Register Window			
RICPIRSI.02	RTCVALH	RTCVALL		
00	MINUTES	SECONDS		
01	WEEKDAY	HOURS		
10	MONTH	DAY		
11	—	YEAR		

The Alarm Value register window (ALRMVALH and ALRMVALL) uses the ALRMPTR bits (ALRMCFG<1:0>) to select the desired Alarm register pair.

By reading or writing to the ALRMVALH register, the Alarm Pointer value, ALRMPTR<1:0>, decrements by '1' until it reaches '00'. Once it reaches '00', the ALRMMIN and ALRMSEC value will be accessible through ALRMVALH and ALRMVALL until the pointer value is manually changed.

TABLE 15-4:ALRMVAL REGISTER
MAPPING

ALRMPTR<1:0>	Alarm Value Register Window			
ALRIVIPIR 1.02	ALRMVALH	ALRMVALL		
0 0	ALRMMIN	ALRMSEC		
01	ALRMWD	ALRMHR		
10	ALRMMNTH	ALRMDAY		
11	_	_		

15.2.9 CALIBRATION

The real-time crystal input can be calibrated using the periodic auto-adjust feature. When properly calibrated, the RTCC can provide an error of less than three seconds per month.

To perform this calibration, find the number of error clock pulses and store the value into the lower half of the RTCCAL register. The 8-bit, signed value, loaded into RTCCAL, is multiplied by 4 and will either be added or subtracted from the RTCC timer, once every minute.

To calibrate the RTCC module:

- 1. Use another timer resource on the device to find the error of the 32.768 kHz crystal.
- 2. Convert the number of error clock pulses per minute (see Equation 15-1).

EQUATION 15-1: CONVERTING ERROR, CLOCK PULSES

(Ideal Frequency (32,758) – Measured Frequency) * 60 = Error Clocks per Minute

- If the oscillator is *faster* than ideal (negative result from step 2), the RCFGCALL register value needs to be negative. This causes the specified number of clock pulses to be subtracted from the timer counter once every minute.
- If the oscillator is *slower* than ideal (positive result from step 2), the RCFGCALL register value needs to be positive. This causes the specified number of clock pulses to be added to the timer counter once every minute.
- 3. Load the RTCCAL register with the correct value.

Writes to the RTCCAL register should occur only when the timer is turned off or immediately after the rising edge of the seconds pulse.

Note:	In determining the crystal's error value, it
	is the user's responsibility to include the
	crystal's initial error from drift due to
	temperature or crystal aging.

15.3 Alarm

The alarm features and characteristics are:

- Configurable from half a second to one year
- Enabled using the ALRMEN bit (ALRMCFG<7>, Register 15-4)
- · Offers one-time and repeat alarm options

15.3.1 CONFIGURING THE ALARM

The alarm feature is enabled using the ALRMEN bit.

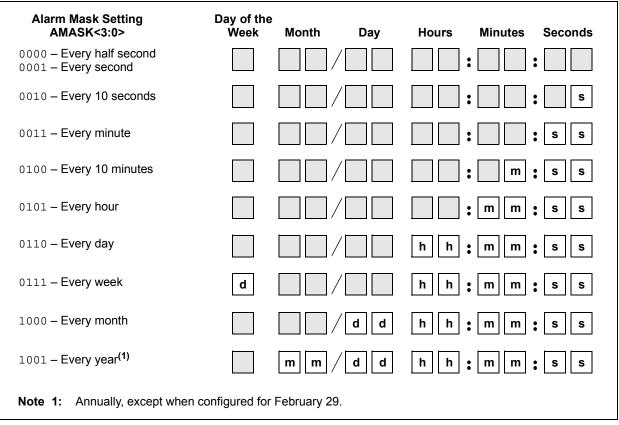
This bit is cleared when an alarm is issued. The bit will not be cleared if the CHIME bit = 1 or if ALRMRPT $\neq 0$.

The interval selection of the alarm is configured through the ALRMCFG (AMASK<3:0>) bits (see Figure 15-5). These bits determine which and how many digits of the alarm must match the clock value for the alarm to occur.

The alarm can also be configured to repeat based on a preconfigured interval. The number of times this occurs, after the alarm is enabled, is stored in the ALRMRPT register.

Note: While the alarm is enabled (ALRMEN = 1), changing any of the registers, other than the RTCCAL, ALRMCFG and ALRMRPT registers and the CHIME bit, can result in a false alarm event leading to a false alarm interrupt. To avoid this, only change the timer and alarm values while the alarm is disabled (ALRMEN = 0). It is recommended that the ALRMCFG and ALRMRPT registers and CHIME bit be changed when RTCSYNC = 0.





When ALRMCFG = 00 and the CHIME bit = 0 (ALRMCFG<6>), the repeat function is disabled and only a single alarm will occur. The alarm can be repeated up to 255 times by loading the ALRMRPT register with FFh.

After each alarm is issued, the ALRMRPT register is decremented by one. Once the register has reached '00', the alarm will be issued one last time.

After the alarm is issued a last time, the ALRMEN bit is cleared automatically and the alarm turned off. Indefinite repetition of the alarm can occur if the CHIME bit = 1.

When CHIME = 1, the alarm is not disabled when the ALRMRPT register reaches '00', but it rolls over to FF and continues counting indefinitely.

15.3.2 ALARM INTERRUPT

At every alarm event, an interrupt is generated. Additionally, an alarm pulse output is provided that operates at half the frequency of the alarm.

The alarm pulse output is completely synchronous with the RTCC clock and can be used as a trigger clock to other peripherals. This output is available on the RTCC pin. The output pulse is a clock with a 50% duty cycle and a frequency, half that of the alarm event (see Figure 15-6).

The RTCC pin also can output the seconds clock. The user can select between the alarm pulse, generated by the RTCC module, or the seconds clock output.

The RTSECSEL<1:0> (PADCFG1<2:1>) bits select between these two outputs:

- Alarm Pulse RTSECSEL<1:0> = 00
- Seconds Clock RTSECSEL<1:0> = 01

FIGURE 15-6:	TIMER PULSE GENERATION
RTCEN b	it
ALRMEN b	it
RTCC Alarm Ever	
RTCC Pi	

15.4 Sleep Mode

The timer and alarm continue to operate while in Sleep mode. The operation of the alarm is not affected by Sleep, as an alarm event can always wake-up the CPU.

The Idle mode does not affect the operation of the timer or alarm.

15.5 Reset

15.5.1 DEVICE RESET

When a device Reset occurs, the ALCFGRPT register is forced to its Reset state causing the alarm to be disabled (if enabled prior to the Reset). If the RTCC was enabled, it will continue to operate when a basic device Reset occurs.

15.5.2 POWER-ON RESET (POR)

The RTCCFG and ALRMRPT registers are reset only on a POR. Once the device exits the POR state, the clock registers should be reloaded with the desired values.

The timer prescaler values can be reset only by writing to the SECONDS register. No device Reset can affect the prescalers.

15.6 Register Maps

Table 15-5, Table 15-6 and Table 15-7 summarize the registers associated with the RTCC module.

File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	All Resets on Page
RTCCFG	RTCEN		RTCWREN	RTCSYNC	HALFSEC	RTCOE	RTCPTR1	RTCPTR0	64
RTCCAL	CAL7	CAL6	CAL5	CAL4	CAL3	CAL2	CAL1	CAL0	64
PADCFG1		_	_	-	_	RTSECSEL1	RTSECSEL0	—	64
ALRMCFG	ALRMEN	CHIME	AMASK3	AMASK2	AMASK1	AMASK0	ALRMPTR1	ALRMPTR0	64
ALRMRPT	ARPT7	ARPT6	ARPT5	ARPT4	ARPT3	ARPT2	ARPT1	ARPT0	64

TABLE 15-5: RTCC CONTROL REGISTERS

Legend: — = unimplemented, read as '0'. Reset values are shown in hexadecimal for 80-pin devices.

TABLE 15-6: RTCC VALUE REGISTERS

File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	All Resets on Page
RTCVALH	RTCC Value	RTCC Value High Register Window Based on RTCPTR<1:0>							64
RTCVALL	RTCC Value Low Register Window Based on RTCPTR<1:0>							64	

Legend: — = unimplemented, read as '0'. Reset values are shown in hexadecimal for 80-pin devices.

TABLE 15-7: ALARM VALUE REGISTERS

File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	All Resets on Page
ALRMVALH	Alarm Value High Register Window Based on ALRMPTR<1:0>						64		
ALRMVALL	LRMVALL Alarm Value Low Register Window Based on ALRMPTR<1:0>							64	

Legend: — = unimplemented, read as '0'. Reset values are shown in hexadecimal for 80-pin devices.

NOTES:

16.0 CAPTURE/COMPARE/PWM (CCP) MODULES

PIC18F87J90 family devices have two CCP (Capture/Compare/PWM) modules, designated CCP1 and CCP2. Both modules implement standard capture, compare and Pulse-Width Modulation (PWM) modes.

Each CCP module contains two 8-bit registers that can operate as two 8-bit Capture registers, two 8-bit Compare registers or two PWM Master/Slave Duty Cycle registers. For the sake of clarity, all CCP module operation in the following sections is described with respect to CCP2, but is equally applicable to CCP1.

REGISTER 16-1: CCPxCON: CCPx CONTROL REGISTER (CCP1, CCP2 MODULES)

U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	—	DCxB1	DCxB0	CCPxM3	CCPxM2	CCPxM1	CCPxM0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6 Unimplemented: Read as '0' bit 5-4 DCxB<1:0>: PWM Duty Cycle bit 1 and bit 0 for CCPx Module Capture mode: Unused. Compare mode: Unused. PWM mode: These bits are the two Least Significant bits (bit 1 and bit 0) of the 10-bit PWM duty cycle. The eight Most Significant bits (DCx<9:2>) of the duty cycle are found in CCPRxL. bit 3-0 CCPxM<3:0>: CCPx Module Mode Select bits 0000 = Capture/Compare/PWM disabled (resets CCPx module) 0001 = Reserved 0010 = Compare mode, toggle output on match (CCPxIF bit is set) 0011 = Reserved 0100 = Capture mode, every falling edge 0101 = Capture mode, every rising edge 0110 = Capture mode, every 4th rising edge 0111 = Capture mode, every 16th rising edge 1000 = Compare mode: initialize CCPx pin low; on compare match, force CCPx pin high (CCPxIF bit is set) 1001 = Compare mode: initialize CCPx pin high; on compare match, force CCPx pin low (CCPxIF bit is set) 1010 = Compare mode: generate software interrupt on compare match (CCPxIF bit is set, CCPx pin reflects I/O state) 1011 = Compare mode: Special Event Trigger; reset timer; start A/D conversion on CCPx match (CCPxIF bit is set)⁽¹⁾ 11xx = PWM mode

Note 1: CCPxM<3:0> = 1011 will only reset the timer and not start an A/D conversion on a CCP1 match.

16.1 CCP Module Configuration

Each Capture/Compare/PWM module is associated with a control register (generically, CCPxCON) and a data register (CCPRx). The data register, in turn, is comprised of two 8-bit registers: CCPRxL (low byte) and CCPRxH (high byte). All registers are both readable and writable.

16.1.1 CCP MODULES AND TIMER RESOURCES

The CCP modules utilize timers 1, 2 or 3, depending on the mode selected. Timer1 and Timer3 are available to modules in Capture or Compare modes, while Timer2 is available for modules in PWM mode.

TABLE 16-1:CCP MODE – TIMER
RESOURCE

CCP Mode	Timer Resource
Capture	Timer1 or Timer3
Compare	Timer1 or Timer3
PWM	Timer2

The assignment of a particular timer to a module is determined by the Timer to CCP enable bits in the T3CON register (Register 14-1). Both modules may be active at any given time and may share the same timer resource if they are configured to operate in the same mode (Capture/Compare or PWM) at the same time. The interactions between the two modules are summarized in Table 16-2.

Depending on the configuration selected, up to four timers may be active at once, with modules in the same configuration (Capture/Compare or PWM) sharing timer resources. The possible configurations are shown in Figure 16-1.

16.1.2 OPEN-DRAIN OUTPUT OPTION

When operating in Output mode (i.e., in Compare or PWM modes), the drivers for the CCPx pins can be optionally configured as open-drain outputs. This feature allows the voltage level on the pin to be pulled to a higher level through an external pull-up resistor and allows the output to communicate with external circuits without the need for additional level shifters.

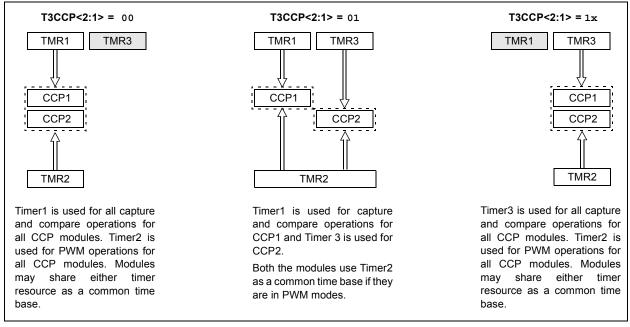
The open-drain output option is controlled by the CCP2OD and CCP1OD bits (TRISG<6:5>). Setting the appropriate bit configures the pin for the corresponding module for open-drain operation.

16.1.3 CCP2 PIN ASSIGNMENT

The pin assignment for CCP2 (capture input, compare and PWM output) can change, based on device configuration. The CCP2MX Configuration bit determines which pin CCP2 is multiplexed to. By default, it is assigned to RC1 (CCP2MX = 1). If the Configuration bit is cleared, CCP2 is multiplexed with RE7.

Changing the pin assignment of CCP2 does not automatically change any requirements for configuring the port pin. Users must always verify that the appropriate TRIS register is configured correctly for CCP2 operation, regardless of where it is located.

FIGURE 16-1: CCP AND TIMER INTERCONNECT CONFIGURATIONS



CCP1 Mode	CCP2 Mode	Interaction
Capture	Capture	Each module can use TMR1 or TMR3 as the time base. The time base can be different for each CCP.
Capture	Compare	CCP2 can be configured for the Special Event Trigger to reset TMR1 or TMR3 (depending upon which time base is used). Automatic A/D conversions on a trigger event can also be done. Operation of CCP1 could be affected if it is using the same timer as a time base.
Compare	Capture	CCP1 can be configured for the Special Event Trigger to reset TMR1 or TMR3 (depending upon which time base is used). Operation of CCP2 could be affected if it is using the same timer as a time base.
Compare	Compare	Either module can be configured for the Special Event Trigger to reset the time base. Automatic A/D conversions on a CCP2 trigger event can be done. Conflicts may occur if both modules are using the same time base.
Capture	PWM	None
Compare	PWM	None
PWM	Capture	None
PWM	Compare	None
PWM	PWM	Both PWMs will have the same frequency and update rate (TMR2 interrupt).

TABLE 16-2: INTERACTIONS BETWEEN CCP1 AND CCP2 FOR TIMER RESOURCES

16.2 Capture Mode

In Capture mode, the CCPR2H:CCPR2L register pair captures the 16-bit value of the TMR1 or TMR3 register when an event occurs on the CCP2 pin (RC1 or RE7, depending on device configuration). An event is defined as one of the following:

- Every falling edge
- Every rising edge
- · Every 4th rising edge
- · Every 16th rising edge

The event is selected by the mode select bits, CCP2M<3:0> (CCP2CON<3:0>). When a capture is made, the interrupt request flag bit, CCP2IF (PIR3<2>), is set; it must be cleared in software. If another capture occurs before the value in register, CCPR2, is read, the old captured value is overwritten by the new captured value.

16.2.1 CCP PIN CONFIGURATION

In Capture mode, the appropriate CCPx pin should be configured as an input by setting the corresponding TRIS direction bit.

Note:	If RC1/CCP2 or RE7/CCP2 is configured
	as an output, a write to the port can cause
	a capture condition.

16.2.2 TIMER1/TIMER3 MODE SELECTION

The timers that are to be used with the capture feature (Timer1 and/or Timer3) must be running in Timer mode or Synchronized Counter mode. In Asynchronous Counter mode, the capture operation may not work. The timer to be used with each CCP module is selected in the T3CON register (see Section 16.1.1 "CCP Modules and Timer Resources").

16.2.3 SOFTWARE INTERRUPT

When the Capture mode is changed, a false capture interrupt may be generated. The user should keep the CCP2IE bit (PIE3<2>) clear to avoid false interrupts and should clear the flag bit, CCP2IF, following any such change in operating mode.

16.2.4 CCP PRESCALER

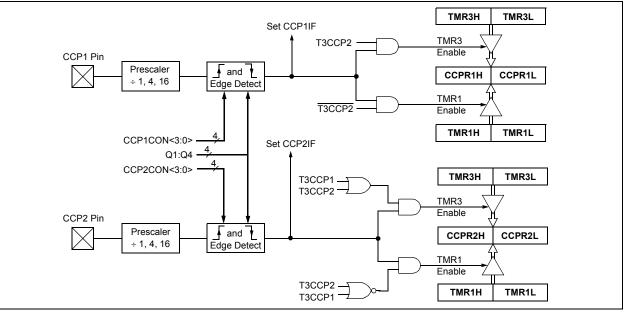
There are four prescaler settings in Capture mode. They are specified as part of the operating mode selected by the mode select bits (CCP2M<3:0>). Whenever the CCP module is turned off, or the CCP module is not in Capture mode, the prescaler counter is cleared. This means that any Reset will clear the prescaler counter.

Switching from one capture prescaler to another may generate an interrupt. Also, the prescaler counter will not be cleared; therefore, the first capture may be from a non-zero prescaler. Example 16-1 shows the recommended method for switching between capture prescalers. This example also clears the prescaler counter and will not generate the "false" interrupt.

EXAMPLE 16-1: CHANGING BETWEEN CAPTURE PRESCALERS

; Turn CCP module off
; Load WREG with the
; new prescaler mode
; value and CCP ON
; Load CCP2CON with
; this value

FIGURE 16-2: CAPTURE MODE OPERATION BLOCK DIAGRAM



16.3 Compare Mode

In Compare mode, the 16-bit CCPR2 register value is constantly compared against either the TMR1 or TMR3 register pair value. When a match occurs, the CCP2 pin can be:

- driven high
- · driven low
- toggled (high-to-low or low-to-high)
- remain unchanged (that is, reflects the state of the I/O latch)

The action on the pin is based on the value of the mode select bits (CCP2M<3:0>). At the same time, the interrupt flag bit, CCP2IF, is set.

16.3.1 CCP PIN CONFIGURATION

The user must configure the CCPx pin as an output by clearing the appropriate TRIS bit.

Clearing the CCP2CON register will force
the RC1 or RE7 compare output latch
(depending on device configuration) to the
default low level. This is not the PORTC or
PORTE I/O data latch.

16.3.2 TIMER1/TIMER3 MODE SELECTION

Timer1 and/or Timer3 must be running in Timer mode, or Synchronized Counter mode, if the CCP module is using the compare feature. In Asynchronous Counter mode, the compare operation may not work.

16.3.3 SOFTWARE INTERRUPT MODE

When the Generate Software Interrupt mode is chosen (CCP2M<3:0> = 1010), the CCP2 pin is not affected. Only a CCP interrupt is generated, if enabled, and the CCP2IE bit is set.

16.3.4 SPECIAL EVENT TRIGGER

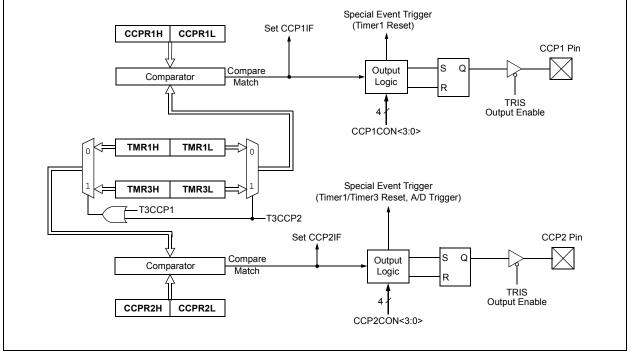
Both CCP modules are equipped with a Special Event Trigger. This is an internal hardware signal generated in Compare mode to trigger actions by other modules. The Special Event Trigger is enabled by selecting the Compare Special Event Trigger mode (CCP2M<3:0> = 1011).

For either CCP module, the Special Event Trigger resets the Timer register pair for whichever timer resource is currently assigned as the module's time base. This allows the CCPRx registers to serve as a programmable period register for either timer.

The Special Event Trigger for CCP2 can also start an A/D conversion. In order to do this, the A/D Converter must already be enabled.

Note: The Special Event Trigger of CCP1 only resets Timer1/Timer3 and cannot start an A/D conversion, even when the A/D Converter is enabled.

FIGURE 16-3: COMPARE MODE OPERATION BLOCK DIAGRAM



					,		·		
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59
RCON	IPEN	—	CM	RI	TO	PD	POR	BOR	60
PIR3	_	LCDIF	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF	62
PIE3	_	LCDIE	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE	62
IPR3	_	LCDIP	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP	62
PIR2	OSCFIF	CMIF	_	—	BCLIF	LVDIF	TMR3IF	—	62
PIE2	OSCFIE	CMIE	—	_	BCLIE	LVDIE	TMR3IE	_	62
IPR2	OSCFIP	CMIP	_	_	BCLIP	LVDIP	TMR3IP	_	62
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	62
TRISE	TRISE7	TRISE6	TRISE5	TRISE4	TRISE3		TRISE1	TRISE0	62
TRISG	SPIOD	CCP2OD	CCP10D	TRISG4	TRISG3	TRISG2	TRISG1	TRISG0	62
TMR1L	Timer1 Register Low Byte								60
TMR1H	Timer1 Register High Byte							60	
T1CON	RD16	T1RUN	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N	60
TMR3H	Timer3 Register High Byte							61	
TMR3L	Timer3 Register Low Byte								61
T3CON	RD16	T3CCP2	T3CKPS1	T3CKPS0	T3CCP1	T3SYNC	TMR3CS	TMR3ON	61
CCPR1L	Capture/Compare/PWM Register 1 Low Byte								63
CCPR1H	Capture/Compare/PWM Register 1 High Byte								63
CCP1CON	—		DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	63
CCPR2L	Capture/Compare/PWM Register 2 Low Byte								64
CCPR2H	Capture/Compare/PWM Register 2 High Byte								63
CCP2CON	_	_	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	64
						0			-

TABLE 16-3: REGISTERS ASSOCIATED WITH CAPTURE, COMPARE, TIMER1 AND TIMER3

Legend: — = unimplemented, read as '0'. Shaded cells are not used by Capture/Compare, Timer1 or Timer3.

16.4 PWM Mode

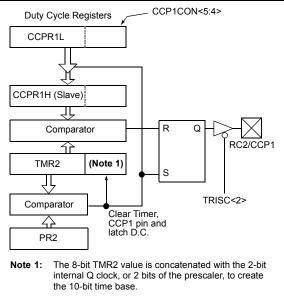
In Pulse-Width Modulation (PWM) mode, the CCP2 pin produces up to a 10-bit resolution PWM output. Since the CCP2 pin is multiplexed with a PORTC or PORTE data latch, the appropriate TRIS bit must be cleared to make the CCP2 pin an output.

Note:	Clearing the CCP2CON register will force the RC1 or RE7 output latch (depending on the device configuration) to the default								
	low level. This is not the PORTC or PORTE I/O data latch.								

Figure 16-4 shows a simplified block diagram of the CCP1 module in PWM mode.

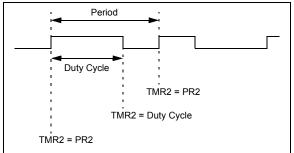
For a step-by-step procedure on how to set up the CCP module for PWM operation, see **Section 16.4.3** "Setup for PWM Operation".





A PWM output (Figure 16-5) has a time base (period) and a time that the output stays high (duty cycle). The frequency of the PWM is the inverse of the period (1/period).

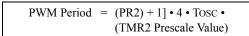
FIGURE 16-5: PWM OUTPUT



16.4.1 PWM PERIOD

The PWM period is specified by writing to the PR2 register. The PWM period can be calculated using the following formula:

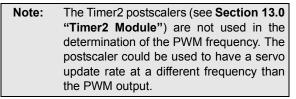
EQUATION 16-1:



PWM frequency is defined as 1/[PWM period].

When TMR2 is equal to PR2, the following three events occur on the next increment cycle:

- TMR2 is cleared
- The CCP2 pin is set (exception: if PWM duty cycle = 0%, the CCP2 pin will not be set)
- The PWM duty cycle is latched from CCPR2L into CCPR2H



16.4.2 PWM DUTY CYCLE

The PWM duty cycle is specified by writing to the CCPR2L register and to the CCP2CON<5:4> bits. Up to 10-bit resolution is available. The CCPR2L contains the eight MSbs and the CCP2CON<5:4> contains the two LSbs. This 10-bit value is represented by CCPR2L:CCP2CON<5:4>. The following equation is used to calculate the PWM duty cycle in time:

EQUATION 16-2:

PWM Duty Cycle = (CCPR2L:CCP2CON<5:4>) • Tosc • (TMR2 Prescale Value)

CCPR2L and CCP2CON<5:4> can be written to at any time, but the duty cycle value is not latched into CCPR2H until after a match between PR2 and TMR2 occurs (i.e., the period is complete). In PWM mode, CCPR2H is a read-only register.

The CCPR2H register and a 2-bit internal latch are used to double-buffer the PWM duty cycle. This double-buffering is essential for glitchless PWM operation.

When the CCPR2H and 2-bit latch match TMR2, concatenated with an internal 2-bit Q clock or 2 bits of the TMR2 prescaler, the CCP2 pin is cleared.

The maximum PWM resolution (bits) for a given PWM frequency is given by the equation:

EQUATION 16-3:

PWM Resolution (max) =
$$\frac{\log(\frac{FOSC}{FPWM})}{\log(2)}$$
 bits

Note: If the PWM duty cycle value is longer than the PWM period, the CCP2 pin will not be cleared.

TABLE 16-4: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS AT 40 MHz

PWM Frequency	2.44 kHz	9.77 kHz	39.06 kHz	156.25 kHz	312.50 kHz	416.67 kHz
Timer Prescaler (1, 4, 16)	16	4	1	1	1	1
PR2 Value	FFh	FFh	FFh	3Fh	1Fh	17h
Maximum Resolution (bits)	14	12	10	8	7	6.58

16.4.3 SETUP FOR PWM OPERATION

The following steps should be taken when configuring the CCP module for PWM operation:

- 1. Set the PWM period by writing to the PR2 register.
- 2. Set the PWM duty cycle by writing to the CCPR2L register and CCP2CON<5:4> bits.
- Make the CCP2 pin an output by clearing the appropriate TRIS bit.
- 4. Set the TMR2 prescale value, then enable Timer2 by writing to T2CON.
- 5. Configure the CCP2 module for PWM operation.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59
RCON	IPEN	_	CM	RI	TO	PD	POR	BOR	60
PIR1	—	ADIF	RC1IF	TX1IF	SSPIF	_	TMR2IF	TMR1IF	62
PIE1	_	ADIE	RC1IE	TX1IE	SSPIE		TMR2IE	TMR1IE	62
IPR1	_	ADIP	RC1IP	TX1IP	SSPIP	-	TMR2IP	TMR1IP	62
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	62
TRISE	TRISE7	TRISE6	TRISE5	TRISE4	TRISE3		TRISE1	TRISE0	62
TRISG	SPIOD	CCP2OD	CCP10D	TRISG4	TRISG3	TRISG2	TRISG1	TRISG0	62
TMR2	Timer2 Reg	gister							60
PR2	Timer2 Per	iod Register							60
T2CON	_	T2OUTPS3	T2OUTPS2	T2OUTPS1	T2OUTPS0	TMR2ON	T2CKPS1	T2CKPS0	60
CCPR1L	Capture/Co	mpare/PWN	I Register 1 I	_ow Byte					63
CCPR1H	Capture/Co	mpare/PWN	I Register 1 I	High Byte					63
CCP1CON	_	_	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	63
CCPR2L	Capture/Compare/PWM Register 2 Low Byte							64	
CCPR2H	Capture/Co	mpare/PWN	I Register 2 I	High Byte					63
CCP2CON			DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	64

TABLE 16-5: REGISTERS ASSOCIATED WITH PWM AND TIMER2

Legend: — = unimplemented, read as '0'. Shaded cells are not used by PWM or Timer2.

NOTES:

17.0 LIQUID CRYSTAL DISPLAY (LCD) DRIVER MODULE

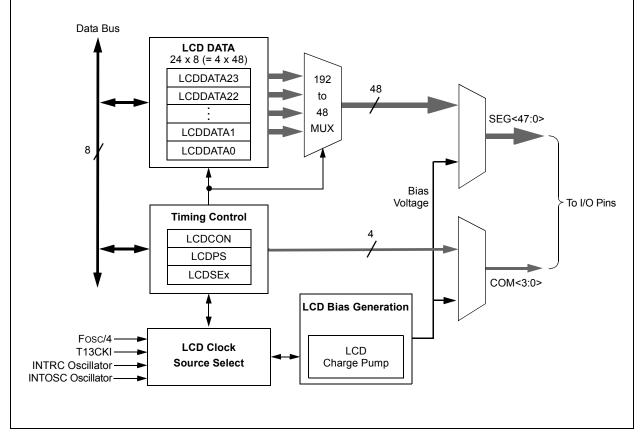
The Liquid Crystal Display (LCD) driver module generates the timing control to drive a static or multiplexed LCD panel. It also provides control of the LCD pixel data. The module can drive panels of up to 192 pixels (48 segments by 4 commons) in PIC18F8XJ90 devices and 132 pixels (33 segments by 4 commons) in PIC18F6XJ90 devices.

The LCD driver module supports these features:

- · Direct driving of LCD panel
- On-chip bias generator with dedicated charge pump to support a range of fixed and variable bias options
- Up to four commons, with four Multiplexing modes
- Up to 48 (PIC18F8XJ90 devices) or 33 (PIC18F6XJ90 devices) segments
- Three LCD clock sources with selectable prescaler, with a fourth source available for use with the LCD charge pump

A simplified block diagram of the module is shown in Figure 17-1.





17.1 LCD Registers

The LCD driver module has 33 registers:

- LCD Control Register (LCDCON)
- LCD Phase Register (LCDPS)
- LCDREG Register (LCD Regulator Control)
- Six LCD Segment Enable Registers (LCDSE5:LCDSE0)
- 24 LCD Data Registers (LCDDATA23:LCDDATA0)

17.1.1 LCD CONTROL REGISTERS

The LCDCON register, shown in Register 17-1, controls the overall operation of the module. Once the module is configured, the LCDEN (LCDCON<7>) bit is used to enable or disable the LCD module. The LCD panel can also operate during Sleep by clearing the SLPEN (LCDCON<6>) bit.

The LCDPS register, shown in Register 17-2, configures the LCD clock source prescaler and the type of waveform: Type-A or Type-B. Details on these features are provided in Section 17.2 "LCD Clock Source", Section 17.3 "LCD Bias Generation" and Section 17.8 "LCD Waveform Generation".

The LCDREG register is described in Section 17.3 "LCD Bias Generation".

The LCD Segment Enable registers (LCDSEx) configure the functions of the port pins. Setting the segment enable bit for a particular segment configures that pin as an LCD driver. The prototype LCDSE register is shown in Register 17-3. There are six LCDSE registers (LCDSE5:LCDSE0) listed in Table 17-1.

REGISTER 17-1: LCDCON: LCD CONTROL REGISTER

bit 1-0		(Timer1) clock (Fosc/4) Commons Sele Multiplex		Maximum Nur	nber of Pixels:	Bia	з Туре								
bit 1-0	00 = System	clock (Fosc/4)	ect bits												
		· /													
	•	,													
							1x = INTRC (31 kHz)								
bit 3-2	•	ck Source Sele													
bit 4	Unimplemen	ted: Read as '0)'												
bit 5		•)PS<4> = 0 (mus	st be cleared in s	software)									
Site	1 = LCD drive	er module is dis er module is ena	abled in Slee	ep mode											
bit 6		r module is dis Driver Enable i		te hit											
	1 = LCD drive	r module is ena	abled												
bit 7	LCDEN: LCD	Driver Enable	bit												
-n = Value at F	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	iown								
R = Readable	bit	W = Writable b	oit	U = Unimplen	nented bit, read	as '0'									
Legend:		C = Clearable	bit												
bit 7							bit 0								
	SLPEN	WERR		CS1	CS0	LMUX1	LMUX0								
R/W-0	R/W-0	R/C-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0								

LMUX<1:0>	Multiplay Type	Maximum Nur	nber of Pixels:	Pigo Turo
	Multiplex Type	PIC18F6XJ90	PIC18F8XJ90	Bias Type
00	Static (COM0)	33	48	Static
01	1/2 (COM1:COM0)	66	96	1/2 or 1/3
10	1/3 (COM2:COM0)	99	144	1/2 or 1/3
11	1/4 (COM3:COM0)	132	192	1/3

R/W-0	R/W-0	R-0	R-0	R/W-0	R/W-0	R/W-0	R/W-0			
WFT	BIASMD	LCDA	WA	LP3	LP2	LP1	LP0			
bit 7				·			bit			
Legend:										
R = Readab	ole bit	W = Writable	bit	U = Unimplem	nented bit, read	d as '0'				
-n = Value a	It POR	'1' = Bit is set		'0' = Bit is clea		x = Bit is unkr	nown			
bit 7	WET: Wayefo	orm Type Select	bit							
				each frame bou	ndarv)					
				hin each commo						
bit 6	BIASMD: Bia	s Mode Select	bit							
	When LMUX	<1:0> = <u>00:</u>								
	0 = Static Bia	s mode (do not	set this bit to	'1')						
		< <u>1:0> = 01 or 1</u>	<u>0:</u>							
	1 = 1/2 Bias mode 0 = 1/3 Bias mode									
	0 = 1/3 Blas mode When LMUX<1:0> = 11:									
		node (do not se	et this bit to '1	·')						
bit 5	LCDA: LCD Active Status bit									
	1 = LCD driver module is active									
	0 = LCD drive	er module is ina	ctive							
bit 4	WA: LCD Wr	WA: LCD Write Allow Status bit								
		the LCDDATA								
bit 3-0	LP<3:0>: LC	D Prescaler Se	lect bits							
	1111 = 1:16									
	1110 = 1:15									
	1101 = 1:14 1100 = 1:13									
	1011 = 1:12									
	1010 = 1:11									
	1001 = 1:10									
	1000 = 1:9									
	0111 = 1:8 0110 = 1:7									
	0101 = 1:6									
	0100 = 1 :5									
	0011 = 1:4									
	0010 = 1:3 0001 = 1:2									
	0001 = 1.2 0000 = 1.1									

REGISTER 17-2: LCDPS: LCD PHASE REGISTER

REGISTER 17-3: LCDSEx: LCD SEGMENT ENABLE REGISTERS

R/W-0	R/W-0						
SE(n + 7)	SE(n + 6)	SE(n + 5)	SE(n + 4)	SE(n + 3)	SE(n + 2)	SE(n + 1)	SE(n)
bit 7							bit 0

Legend:				
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown	

bit 7-0 SEG(n + 7):SEG(n): Segment Enable bits For LCDSE0: n = 0 For LCDSE1: n = 8 For LCDSE2: n = 16 For LCDSE3: n = 24 For LCDSE4: n = 32 For LCDSE5: n = 40 1 = Segment function of the pin is enabled, digital I/O disabled 0 = I/O function of the pin is enabled

TABLE 17-1: LCDSE REGISTERS AND ASSOCIATED SEGMENTS

Register	Segments
LCDSE0	7:0
LCDSE1	15:8
LCDSE2	23:16
LCDSE3	31:24
LCDSE4 ⁽¹⁾	39:32
LCDSE5 ⁽²⁾	47:40

Note 1: LCDSE4<7:1> (SEG<39:33>) registers are not implemented in PIC18F6XJ90 devices.

2: LCDSE5 is not implemented in PIC18F6XJ90 devices.

17.1.2 LCD DATA REGISTERS

Once the module is initialized for the LCD panel, the individual bits of the LCDDATA23:LCDDATA0 registers are cleared or set to represent a clear or dark pixel, respectively. Specific sets of LCDDATA registers are used with specific segments and common signals. Each bit represents a unique combination of a specific segment connected to a specific common.

Individual LCDDATA bits are named by the convention "SxxCy", with "xx" as the segment number and "y" as the common number. The relationship is summarized in Table 17-2. The prototype LCDDATA register is shown in Register 17-4.

Note: In 64-pin devices, writing into the registers LCDDATA5, LCDDATA11, LCDDATA17 and LCDDATA23, will not affect the status of any pixels.

REGISTER 17-4: LCDDATAX: LCD DATA REGISTERS

R/W-0	R/W-0						
S(n + 7)Cy	S(n + 6)Cy	S(n + 5)Cy	S(n + 4)Cy	S(n + 3)Cy	S(n + 2)Cy	S(n + 1)Cy	S(n)Cy
bit 7							bit 0

Legend:				
R = Readable bit	dable bit W = Writable bit U = Unimplemented bit, read as '0'			
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown	

bit 7-0	S(n + 7)Cy:S(n)Cy: Pixel On bits
	<u>For LCDDATA0 through LCDDATA5: n = (8x), y = 0</u>
	<u>For LCDDATA6 through LCDDATA11: n = (8(x – 6)), y = 1</u>
	<u>For LCDDATA12 through LCDDATA17: n = (8(x – 12)), y = 2</u>
	<u>For LCDDATA18 through LCDDATA23: n = (8(x – 18)), y = 3</u>
	1 = Pixel on (dark)
	0 = Pixel off (clear)

TABLE 17-2: LCDDATA REGISTERS AND BITS FOR SEGMENT AND COM COMBINATIONS

Sogmonto	COM Lines							
Segments	0	1	2	3				
0 through 7	LCDDATA0	LCDDATA6	LCDDATA12	LCDDATA18				
0 through 7	S00C0:S07C0	S00C1:S07C1	S00C2:S07C2	S00C3:S07C3				
9 through 1E	LCDDATA1	LCDDATA7	LCDDATA13	LCDDATA19				
8 through 15	S08C0:S15C0	S08C1:S15C1	S08C2:S15C2	S08C0:S15C3				
16 through 23	LCDDATA2	LCDDATA8	LCDDATA14	LCDDATA20				
To through 23	S16C0:S23C0	S16C1:S23C1	S16C2:S23C2	S16C3:S23C3				
24 through 31	LCDDATA3	LCDDATA9	LCDDATA15	LCDDATA21				
24 through 51	S24C0:S31C0	S24C1:S31C1	S24C2:S31C2	S24C3:S31C3				
22 through 20	LCDDATA4 ⁽¹⁾	LCDDATA10 ⁽¹⁾	LCDDATA16 ⁽¹⁾	LCDDATA22 ⁽¹⁾				
32 through 39	S32C0:S39C0	S32C1:S39C1	S32C2:S39C2	S32C3:S39C3				
40 through 47	LCDDATA5 ⁽²⁾	LCDDATA11 ⁽²⁾	LCDDATA17 ⁽²⁾	LCDDATA23 ⁽²⁾				
40 tillough 47	S40C0:S47C0	S40C1:S47C1	S40C2:S47C2	S40C3:S47C3				

Note 1: Bits<7:1> of these registers are not implemented in PIC18F6XJ90 devices. Bit 0 of these registers (SEG32Cy) is always implemented.

2: These registers are not implemented on PIC18F6XJ90 devices.

17.2 LCD Clock Source

The LCD driver module generates its internal clock from 3 possible sources:

- System clock (Fosc/4)
- · Timer1 oscillator
- INTRC source

The LCD clock generator uses a configurable divide-by-32/divide-by-8192 postscaler to produce a baseline frequency of about 1 kHz nominal, regardless of the source selected. The clock source selection and the postscaler configuration are determined by the Clock Source Select bits, CS<1:0> (LCDCON<3:2>).

An additional programmable prescaler is used to derive the LCD frame frequency from the 1 kHz baseline. The prescaler is configured using the LP<3:0> bits (LCDPS<3:0>) for any one of 16 options, ranging from 1:1 to 1:16.

Proper timing for waveform generation is set by the LMUX<1:0> bits (LCDCON<1:0>). These bits determine which Commons Multiplexing mode is to be used and divide down the LCD clock source as required. They also determine the configuration of the ring counter that is used to switch the LCD commons on or off.

17.2.1 LCD VOLTAGE REGULATOR CLOCK SOURCE

In addition to the clock source for LCD timing, a separate 31 kHz nominal clock is required for the LCD charge pump. This is provided from a distinct branch of the LCD clock source.

The charge pump clock can use either the Timer1 oscillator or the INTRC source, as well as the 8 MHz INTOSC source (after being divided by 256 by a prescaler). The charge pump clock source is configured using the CKSEL<1:0> bits (LCDREG<1:0>).

17.2.2 CLOCK SOURCE CONSIDERATIONS

When using the system clock as the LCD clock source, it is assumed that the system clock frequency is a nominal 32 MHz (for a Fosc/4 frequency of 8 MHz). Because the prescaler option for the Fosc/4 clock selection is fixed at divide-by-8192, system clock speeds that differ from 32 MHz will produce frame frequencies and refresh rates different than discussed in this chapter. The user will need to keep this in mind when designing the display application.

The Timer1 and INTRC sources can be used as LCD clock sources when the device is in Sleep mode. To use the Timer1 oscillator, it is necessary to set the T1OSCEN bit (T1CON<3>). Selecting either Timer1 or INTRC as the LCD clock source will not automatically activate these sources.

Similarly, selecting the INTOSC as the charge pump clock source will not turn the oscillator on. To use INTOSC, it must be selected as the system clock source by using the FOSC2 Configuration bit.

If Timer1 is used as a clock source for the device, either as an LCD clock source or for any other purpose, LCD segment 32 become unavailable.

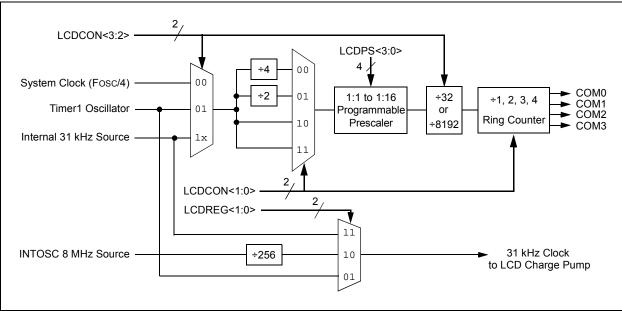


FIGURE 17-2: LCD CLOCK GENERATION

17.3 LCD Bias Generation

The LCD driver module is capable of generating the required bias voltages for LCD operation with a minimum of external components. This includes the ability to generate the different voltage levels required by the different bias types that are required by the LCD. The driver module can also provide bias voltages, both above and below microcontroller VDD, through the use of an on-chip LCD voltage regulator.

17.3.1 LCD BIAS TYPES

PIC18F87J90 family devices support three bias types based on the waveforms generated to control segments and commons:

- Static (two discrete levels)
- 1/2 Bias (three discrete levels
- 1/3 Bias (four discrete levels)

The use of different waveforms in driving the LCD is discussed in more detail in **Section 17.8 "LCD Waveform Generation"**.

17.3.2 LCD VOLTAGE REGULATOR

The purpose of the LCD regulator is to provide proper bias voltage and good contrast for the LCD, regardless of VDD levels. This module contains a charge pump and internal voltage reference. The regulator can be configured by using external components to boost bias voltage above VDD. It can also operate a display at a constant voltage below VDD. The regulator can also be selectively disabled to allow bias voltages to be generated by an external resistor network.

The LCD regulator is controlled through the LCDREG register (Register 17-5). It is enabled or disabled using the CKSEL<1:0> bits, while the charge pump can be selectively enabled using the CPEN bit. When the regulator is enabled, the MODE13 bit is used to select the bias type. The peak LCD bias voltage, measured as a difference between the potentials of LCDBIAS3 and LCDBIAS0, is configured with the BIAS bits.

REGISTER 17-5: LCDREG: VOLTAGE REGULATOR CONTROL REGISTER

U-0	RW-0	RW-1	RW-1	RW-1	RW-1	RW-0	RW-0
—	CPEN	BIAS2	BIAS1	BIAS0	MODE13	CKSEL1	CKSEL0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	Unimplemented: Read as '0'
bit 6	CPEN: LCD Charge Pump Enable bit
	1 = Charge pump enabled; highest LCD bias voltage is 3.6V0 = Charge pump disabled; highest LCD bias voltage is AVDD
bit 5-3	BIAS<2:0>: Regulator Voltage Output Control bits
	111 = 3.60V peak (offset on LCDBIAS0 of 0V)
	110 = 3.47V peak (offset on LCDBIAS0 of 0.13V)
	101 = 3.34V peak (offset on LCDBIAS0 of 0.26V)
	100 = 3.21V peak (offset on LCDBIAS0 of 0.39V)
	011 = 3.08V peak (offset on LCDBIAS0 of 0.52V)
	010 = 2.95V peak (offset on LCDBIAS0 of 0.65V)
	001 = 2.82V peak (offset on LCDBIAS0 of 0.78V)
	000 = 2.69V peak (offset on LCDBIAS0 of 0.91V)
bit 2	MODE13: 1/3 LCD Bias Enable bit
	1 = Regulator output supports 1/3 LCD Bias mode
	0 = Regulator output supports static LCD Bias mode
bit 1-0	CKSEL<1:0>: Regulator Clock Source Select bits
	11 = INTRC
	10 = INTOSC 8 MHz source
	01 = Timer1 oscillator
	00 = LCD regulator disabled

17.3.3 BIAS CONFIGURATIONS

PIC18F87J90 family devices have four distinct circuit configurations for LCD bias generation:

- M0: Regulator with Boost
- M1: Regulator without Boost
- · M2: Resistor Ladder with Software Contrast
- M3: Resistor Ladder with Hardware Contrast

17.3.3.1 M0 (Regulator with Boost)

In M0 operation, the LCD charge pump feature is enabled. This allows the regulator to generate voltages up to +3.6V to the LCD (as measured at LCDBIAS3).

M0 uses a flyback capacitor connected between VLCAP1 and VLCAP2, as well as filter capacitors on LCDBIAS0 through LCDBIAS3, to obtain the required voltage boost (Figure 17-3). The output voltage (VBIAS) is the difference of potential between LCDBIAS3 and LCDBIAS0. It is set by the BIAS<2:0> bits which adjust the offset between LCDBIAS0 and VSs. The flyback capacitor (CFLY) acts as a charge storage element for large LCD loads. This mode is useful in those cases where the voltage requirements of the LCD are higher than the microcontroller's VDD. It also permits software control of the display's contrast, by adjustment of bias voltage, by changing the value of the BIAS bits.

M0 supports Static and 1/3 Bias types. Generation of the voltage levels for 1/3 Bias is handled automatically, but must be configured in software.

M0 is enabled by selecting a valid regulator clock source (CKSEL<1:0> set to any value except '00') and setting the CPEN bit. If Static Bias type is required, the MODE13 bit must be cleared.

17.3.3.2 M1 (Regulator without Boost)

M1 operation is similar to M0, but does not use the LCD charge pump. It can provide VBIAS up to the voltage level supplied directly to LCDBIAS3. It can be used in cases where VDD for the application is expected to never drop below a level that can provide adequate contrast for the LCD. The connection of external components is very similar to M0, except that LCDBIAS3 must be tied directly to VDD (Figure 17-3).

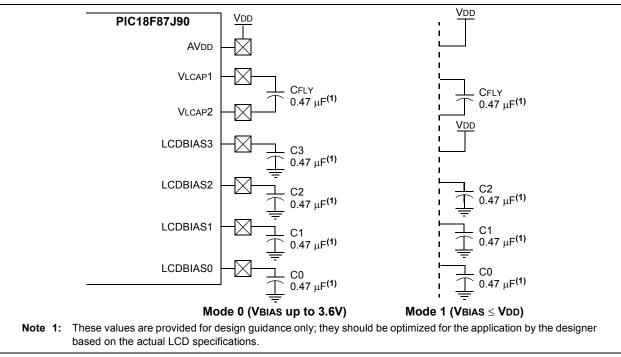
Note:	When the device is put to Sleep while oper-
	ating in mode M0 or M1, make sure that the
	bias capacitors are fully discharged to get
	the lowest Sleep current.

The BIAS<2:0> bits can still be used to adjust contrast in software by changing VBIAS. As with M0, changing these bits changes the offset between LCDBIAS0 and Vss. In M1, this is reflected in the change between the LCDBIAS0 and the voltage tied to LCDBIAS3. Thus, if VDD should change, VBIAS will also change; where in M0, the level of VBIAS is constant.

Like M0, M1 supports Static and 1/3 Bias types. Generation of the voltage levels for 1/3 Bias is handled automatically but must be configured in software.

M1 is enabled by selecting a valid regulator clock source (CKSEL<1:0> set to any value except '00') and clearing the CPEN bit. If 1/3 Bias type is required, the MODE13 bit should also be set.

FIGURE 17-3: LCD REGULATOR CONNECTIONS FOR M0 AND M1 CONFIGURATIONS



17.3.3.3 M2 (Resistor Ladder with Software Contrast)

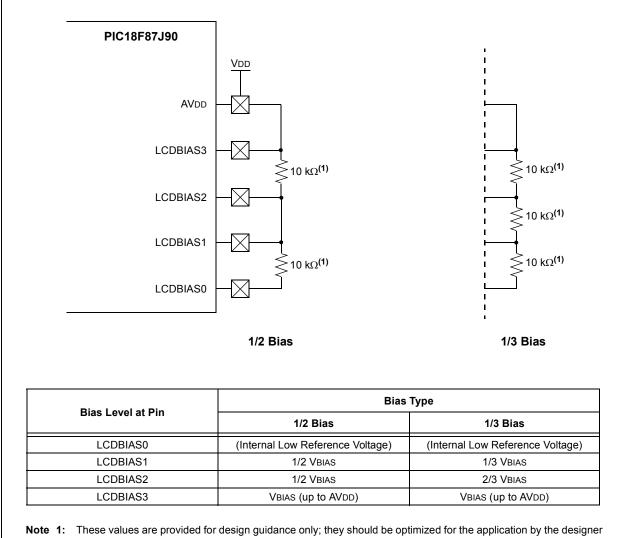
M2 operation also uses the LCD regulator but disables the charge pump. The regulator's internal voltage reference remains active as a way to regulate contrast. It is used in cases where the current requirements of the LCD exceed the capacity of the regulator's charge pump.

In this configuration, the LCD bias voltage levels are created by an external resistor voltage divider, connected across LCDBIAS0 through LCDBIAS3, with the top of the divider tied to VDD (Figure 17-4). The potential at the bottom of the ladder is determined by the LCD regulator's voltage reference, tied internally to LCDBIAS0. The bias type is determined by the voltages on the LCDBIAS pins, which are controlled by the configuration of the resistor ladder. Most applications using M2 will use a 1/3 or 1/2 Bias type. While Static Bias can also be used, it offers extremely limited contrast range and additional current consumption over other bias generation modes.

Like M1, the LCDBIAS bits can be used to control contrast, limited by the level of VDD supplied to the device. Also, since there is no capacitor required across VLCAP1 and VLCAP2, these pins are available as digital I/O ports, RG2 and RG3.

M2 is selected by clearing the CKSEL<1:0> bits and setting the CPEN bit.

FIGURE 17-4: RESISTOR LADDER CONNECTIONS FOR M2 CONFIGURATION



based on the actual LCD specifications.

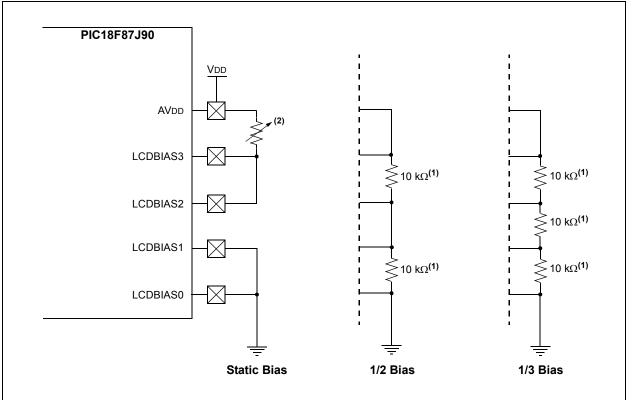
17.3.3.4 M3 (Hardware Contrast)

In M3, the LCD regulator is completely disabled. Like M2, LCD bias levels are tied to AVDD and are generated using an external divider. The difference is that the internal voltage reference is also disabled and the bottom of the ladder is tied to ground (Vss); see Figure 17-5. The value of the resistors, and the difference between Vss and VDD, determine the contrast range; no software adjustment is possible. This configuration is also used where the LCD's current requirements exceed the capacity of the charge pump and software contrast control is not needed.

Depending on the bias type required, resistors are connected between some or all of the pins. A potentiometer can also be connected between LCDBIAS3 and VDD to allow for hardware controlled contrast adjustment.

M3 is selected by clearing the CKSEL<1:0> and CPEN bits.

FIGURE 17-5: RESISTOR LADDER CONNECTIONS FOR M3 CONFIGURATION



Bias Level at Pin		Bias Type	
	Static	1/2 Bias	1/3 Bias
LCDBIAS0	AVss	AVss	AVss
LCDBIAS1	AVss	1/2 AVDD	1/3 AVDD
LCDBIAS2	AVDD	1/2 AVDD	2/3 AVDD
LCDBIAS3	AVDD	AVDD	AVdd

Note 1: These values are provided for design guidance only; they should be optimized for the application by the designer based on the actual LCD specifications.

2: A potentiometer for manual contrast adjustment is optional; it may be omitted entirely.

17.3.4 DESIGN CONSIDERATIONS FOR THE LCD CHARGE PUMP

When designing applications that use the LCD regulator with the charge pump enabled, users must always consider both the dynamic current and RMS (static) current requirements of the display, and what the charge pump can deliver. Both dynamic and static current can be determined by Equation 17-1:

EQUATION 17-1:

$I = C \ge \frac{dV}{dT}$

For dynamic current, C is the value of the capacitors attached to LCDBIAS3 and LCDBIAS2. The variable, dV, is the voltage drop allowed on C2 and C3 during a voltage switch on the LCD display, and dT is the duration of the transient current after a clock pulse occurs. For practical design purposes, these will be assumed to be 0.047 μ F for C, 0.1V for dV and 1 μ s for dT. This yields a dynamic current of 4.7 mA for 1 μ s.

RMS current is determined by the value of CFLY for C, the voltage across VLCAP1 and VLCAP2 for dV and the regulator clock period (TPER) for dT. Assuming a CFLY value of 0.047 μ F, a value of 1.02V across CFLY and TPER of 30 μ s, the maximum theoretical static current will be 1.8 mA. Since the charge pump must charge five capacitors, the maximum current becomes 360 μ A. For a real-world assumption of 50% efficiency, this yields a practical current of 180 μ A.

Users should compare the calculated current capacity against the requirements of the LCD. While dV and dT are relatively fixed by device design, the values of CFLY and the capacitors on the LCDBIAS pins can be changed to increase or decrease current. As always, any changes should be evaluated in the actual circuit for their impact on the application.

17.4 LCD Multiplex Types

The LCD driver module can be configured into four multiplex types:

- Static (only COM0 used)
- 1/2 Multiplex (COM0 and COM1 are used)
- 1/3 Multiplex (COM0, COM1 and COM2 are used)
- 1/4 Multiplex (all COM0, COM1, COM2 and COM3 are used)

The number of active commons used is configured by the LMUX<1:0> bits (LCDCON<1:0>), which determines the function of the PORTE<6:4> pins (see Table 17-3 for details). If the pin is configured as a COM drive, the port I/O function is disabled and the TRIS setting of that pin is overridden.

Note:	On a Power-on	Reset,	the	LMUX<1:0>
	bits are '00'.			

TABLE 17-3: PORTE<6:4> FUNCTION

LMUX<1:0>	PORTE<6>	PORTE<5>	PORTE<4>
0 0	Digital I/O	Digital I/O	Digital I/O
01	Digital I/O	Digital I/O	COM1 Driver
10	Digital I/O	COM2 Driver	COM1 Driver
11	COM3 Driver	COM2 Driver	COM1 Driver

17.5 Segment Enables

The LCDSEx registers are used to select the pin function for each segment pin. Setting a bit configures the corresponding pin to function as a segment driver. LCDSEx registers do not override the TRIS bit settings, so the TRIS bits must be configured as inputs for that pin.

Note: On a Power-on Reset, these pins are configured as digital I/O.

17.6 Pixel Control

The LCDDATAx registers contain bits which define the state of each pixel. Each bit defines one unique pixel.

Table 17-2 shows the correlation of each bit in the LCDDATAx registers to the respective common and segment signals. Any LCD pixel location not being used for display can be used as general purpose RAM.

17.7 LCD Frame Frequency

The rate at which the COM and SEG outputs change is called the LCD frame frequency. Frame frequency is set by the LP<3:0> bits (LCDPS<3:0>) and is also affected by the Multiplex mode being used. The relationship between the Multiplex mode, LP bits setting and frame rate is shown in Table 17-4 and Table 17-5.

TABLE 17-4:	FRAME FREQUENCY
	FORMULAS

Multiplex Mode	Frame Frequency (Hz)				
Static	Clock Source/(4 x 1 x (LP<3:0> + 1))				
1/2	Clock Source/(2 x 2 x (LP<3:0> + 1))				
1/3	Clock Source/(1 x 3 x (LP<3:0> + 1))				
1/4	Clock Source/(1 x 4 x (LP<3:0> + 1))				

TABLE 17-5: APPROXIMATE FRAME FREQUENCY (IN Hz) FOR LP PRESCALER SETTINGS

LP<3:0>	Multiplex Mode					
LP<3:02	Static	1/2	1/3	1/4		
1	125	125	167	125		
2	83	83	111	83		
3	62	62	83	62		
4	50	50	67	50		
5	42	42	56	42		
6	36	36	48	36		
7	31	31	42	31		

17.8 LCD Waveform Generation

LCD waveform generation is based on the principle that the net AC voltage across the dark pixel should be maximized and the net AC voltage across the clear pixel should be minimized. The net DC voltage across any pixel should be zero.

The COM signal represents the time slice for each common, while the SEG contains the pixel data. The pixel signal (COM-SEG) will have no DC component and it can take only one of the two rms values. The higher rms value will create a dark pixel and a lower rms value will create a clear pixel.

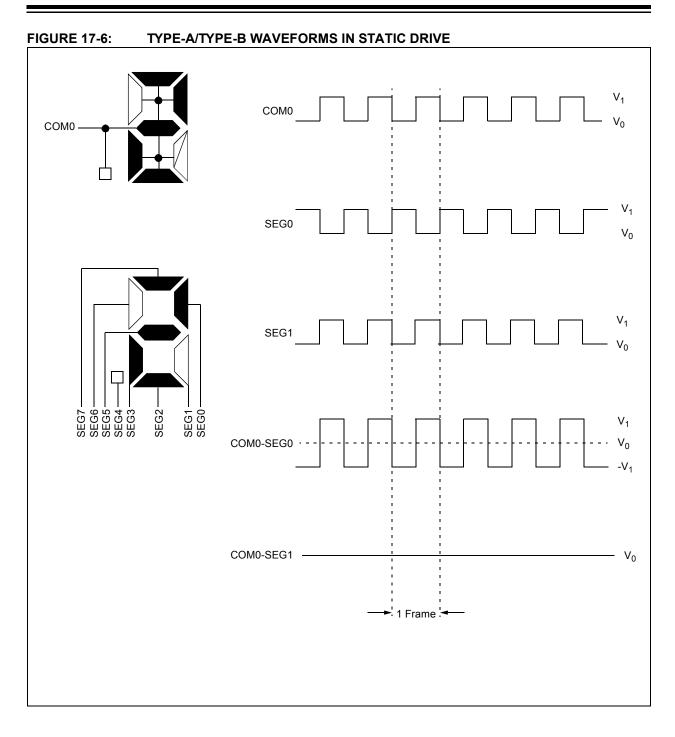
As the number of commons increases, the delta between the two rms values decreases. The delta represents the maximum contrast that the display can have.

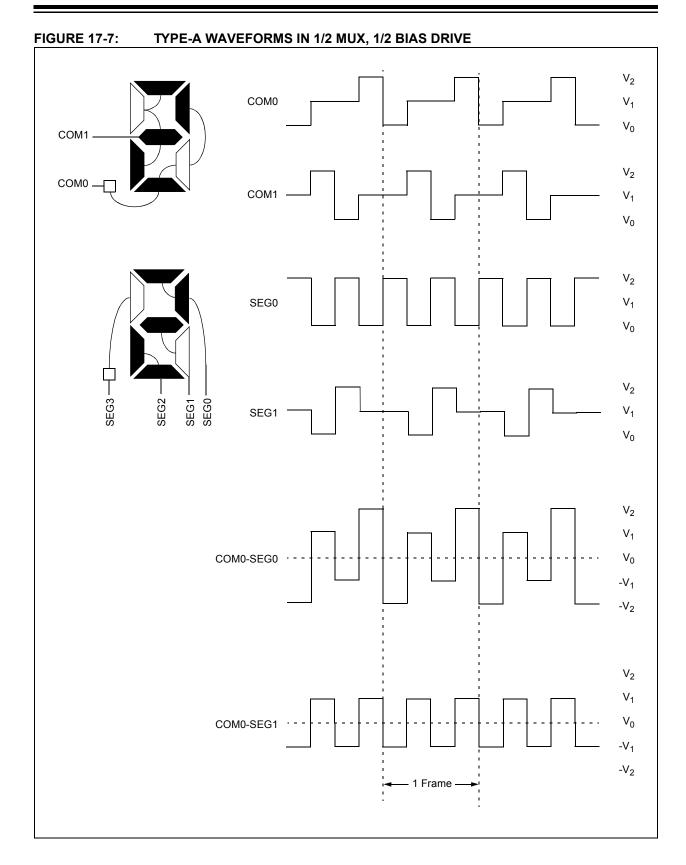
The LCDs can be driven by two types of waveform: Type-A and Type-B. In the Type-A waveform, the phase changes within each common type, whereas in the Type-B waveform, the phase changes on each frame boundary. Thus, the Type-A waveform maintains 0 VDc over a single frame, whereas the Type-B waveform takes two frames.

If the power-managed Sleep mode is
invoked while the LCD Sleep bit (SLPEN)
is set (LCDCON<6> is '1'), take care to
execute Sleep only when the VDC on all
the pixels is '0'.

2: When the LCD clock source is the system clock, the LCD module will go to Sleep if the microcontroller goes into Sleep mode, regardless of the setting of the SLPEN bit. Thus, always take care to see that the VDC on all pixels is '0' whenever Sleep mode is invoked.

Figure 17-6 through Figure 17-16 provide waveforms for static, half multiplex, one-third multiplex and quarter multiplex drives for Type-A and Type-B waveforms.





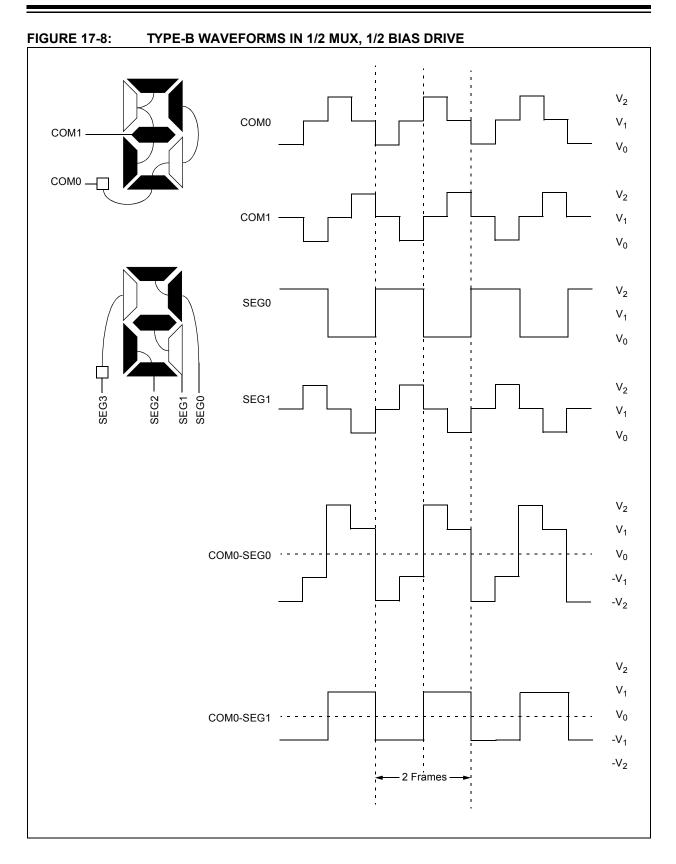
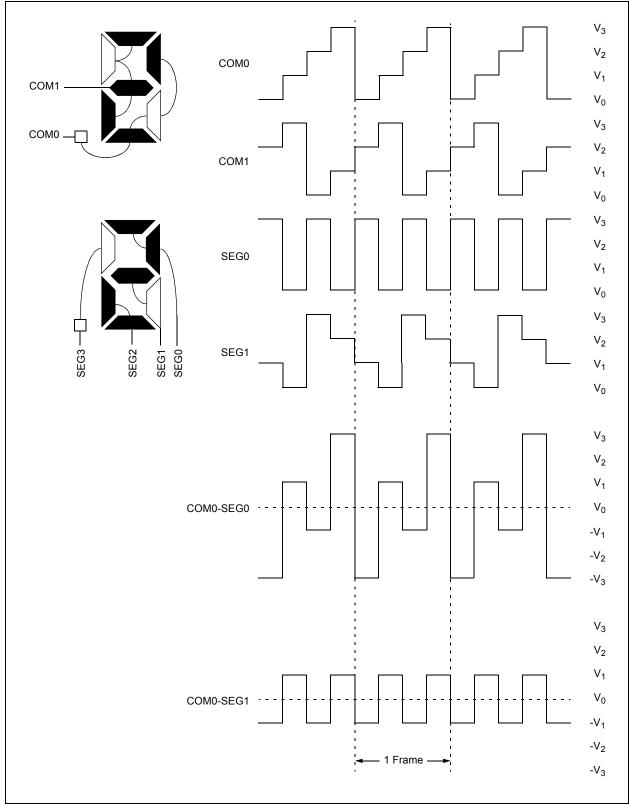
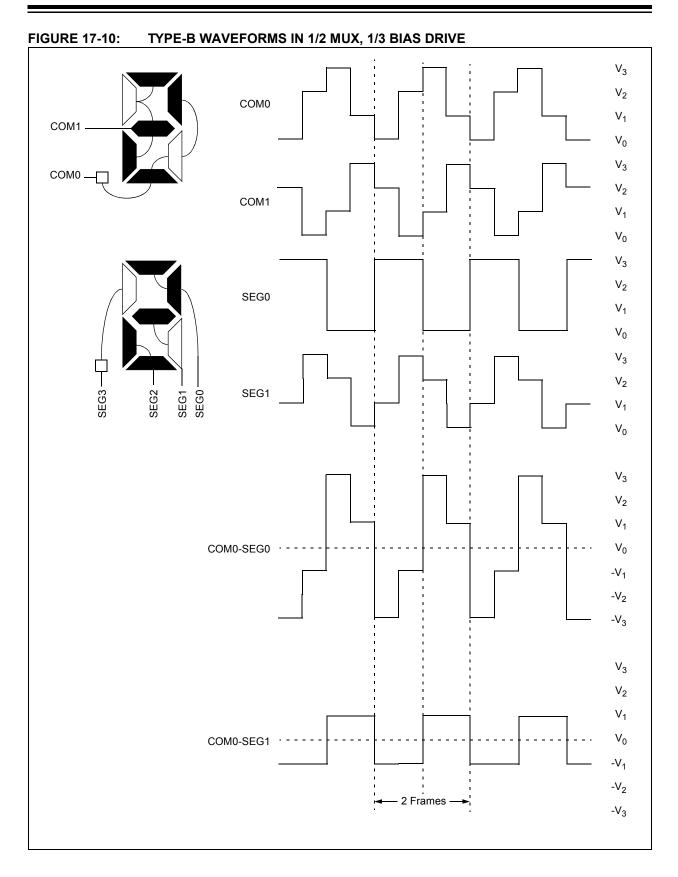


FIGURE 17-9: TYPE-A WAVEFORMS IN 1/2 MUX, 1/3 BIAS DRIVE





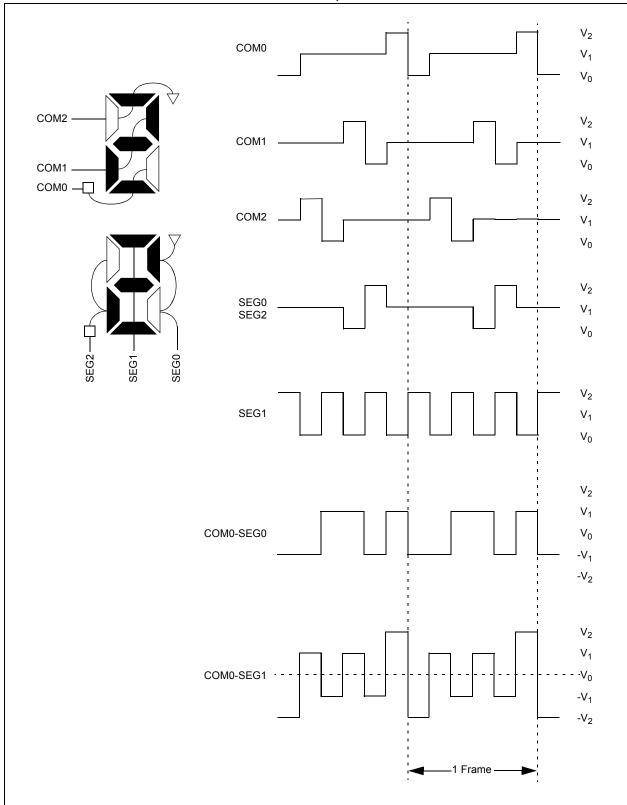
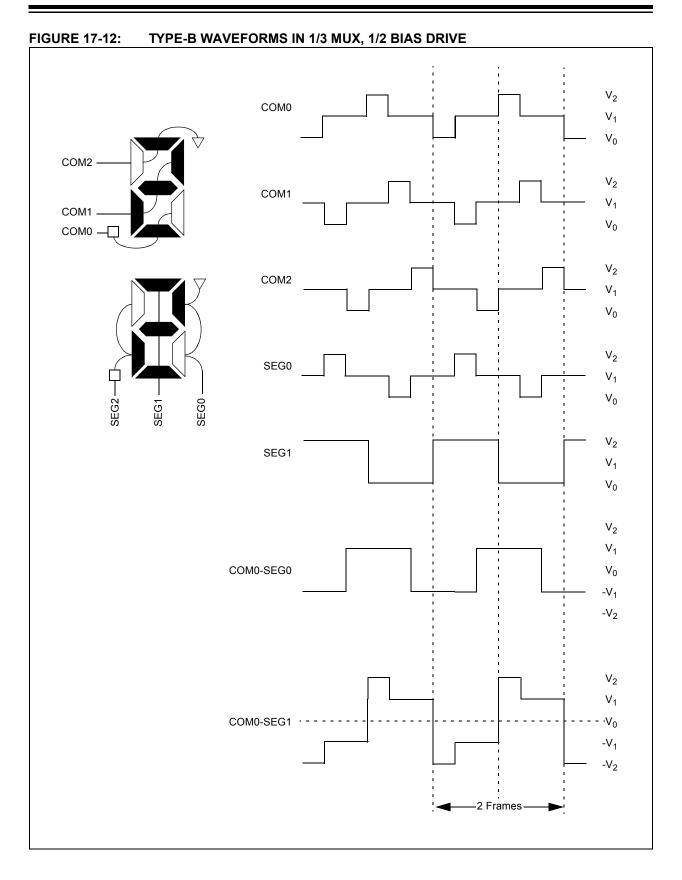


FIGURE 17-11: TYPE-A WAVEFORMS IN 1/3 MUX, 1/2 BIAS DRIVE



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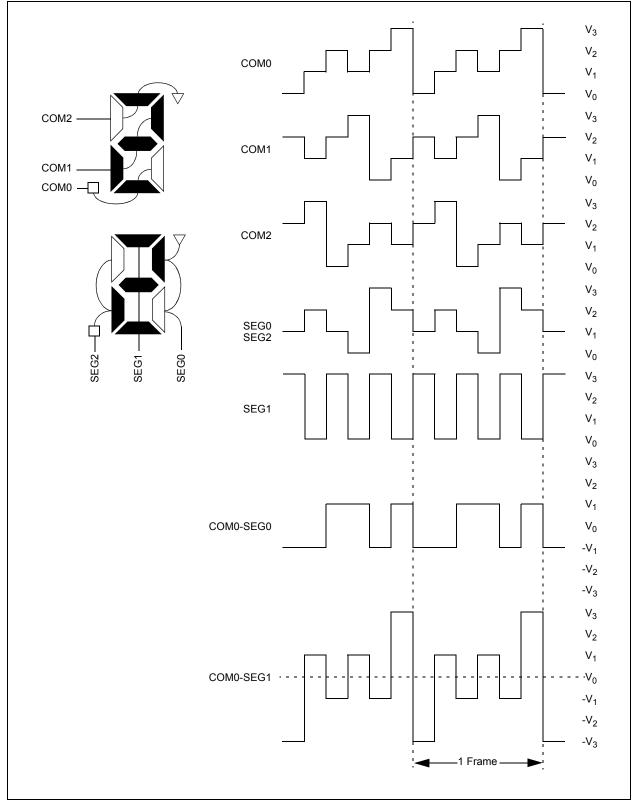
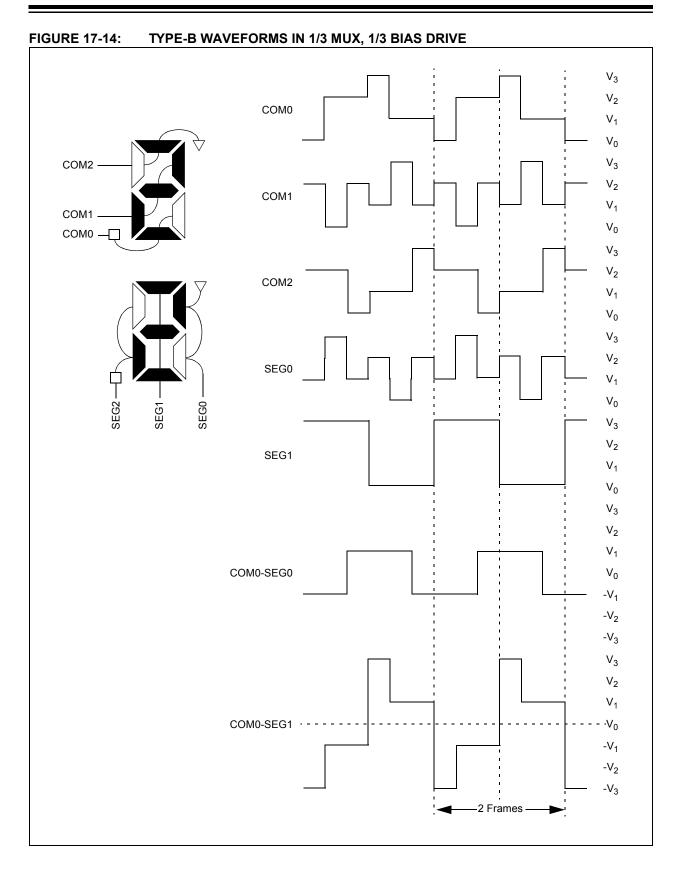
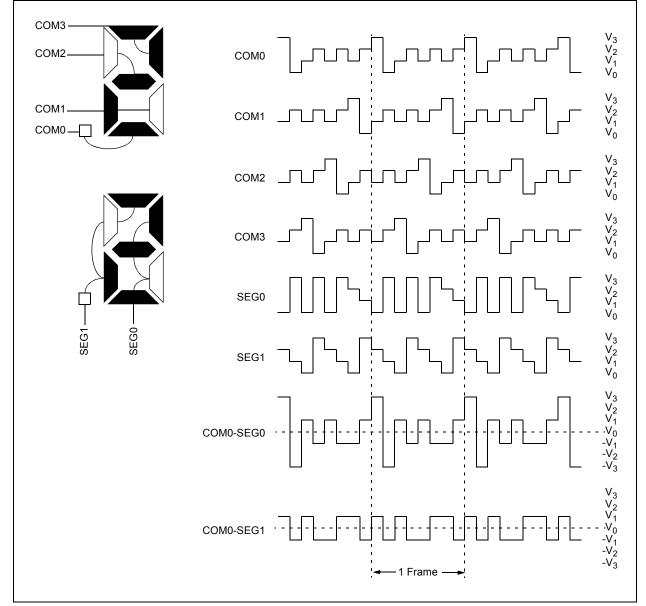
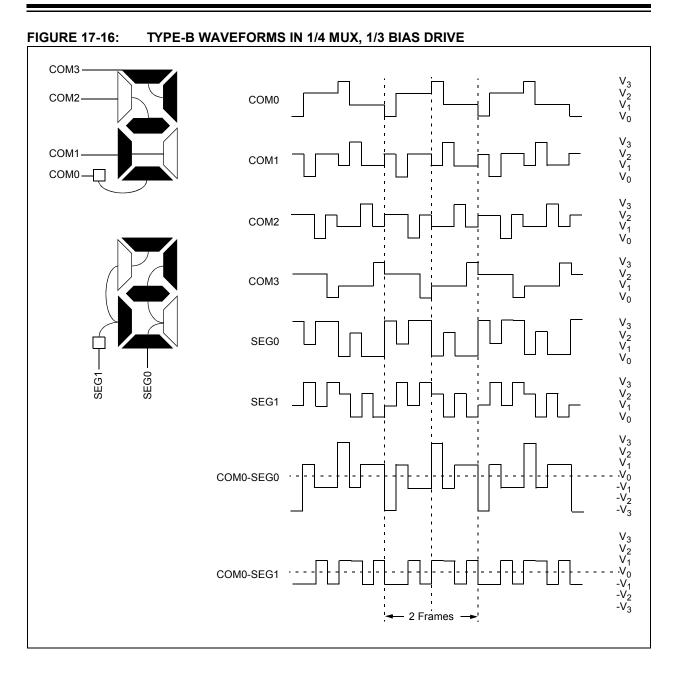


FIGURE 17-13: TYPE-A WAVEFORMS IN 1/3 MUX, 1/3 BIAS DRIVE









17.9 LCD Interrupts

The LCD timing generation provides an interrupt that defines the LCD frame timing. This interrupt can be used to coordinate the writing of the pixel data with the start of a new frame. Writing pixel data at the frame boundary allows a visually crisp transition of the image. This interrupt can also be used to synchronize external events to the LCD. For example, the interface to an external segment driver can be synchronized for a segment data update to the LCD frame.

A new frame is defined to begin at the leading edge of the COM0 common signal. The interrupt will be set immediately after the LCD controller completes accessing all pixel data required for a frame. This will occur at a fixed interval before the frame boundary (TFINT), as shown in Figure 17-17. The LCD controller will begin to access data for the next frame within the interval from the interrupt to when the controller begins to access data after the interrupt (TFWR). New data must be written within TFWR, as this is when the LCD controller will begin to access the data for the next frame. When the LCD driver is running with Type-B waveforms, and the LMUX<1:0> bits are not equal to '00', there are some additional issues that must be addressed. Since the DC voltage on the pixel takes two frames to maintain zero volts, the pixel data must not change between subsequent frames. If the pixel data was allowed to change, the waveform for the odd frames would not necessarily be the complement of the waveform generated in the even frames and a DC component would be introduced into the panel. Therefore, when using Type-B waveforms, the user must synchronize the LCD pixel updates to occur within a subframe after the frame interrupt.

To correctly sequence writing while in Type-B, the interrupt will only occur on complete phase intervals. If the user attempts to write when the write is disabled, the WERR (LCDCON<5>) bit is set.

Note: The interrupt is not generated when the Type-A waveform is selected and when the Type-B with no multiplex (static) is selected.

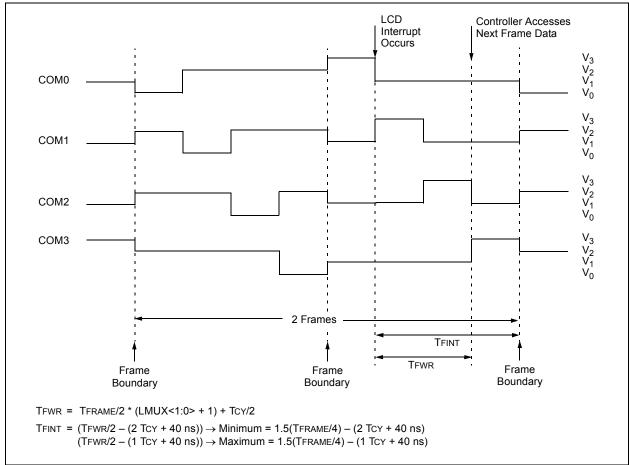


FIGURE 17-17: EXAMPLE WAVEFORMS AND INTERRUPT TIMING IN QUARTER DUTY CYCLE DRIVE

17.10 Operation During Sleep

The LCD module can operate during Sleep. The selection is controlled by the SLPEN bit (LCDCON<6>). Setting the SLPEN bit allows the LCD module to go to Sleep. Clearing the SLPEN bit allows the module to continue to operate during Sleep.

If a SLEEP instruction is executed and SLPEN = 1, the LCD module will cease all functions and go into a very low-current consumption mode. The module will stop operation immediately and drive the minimum LCD voltage on both segment and common lines. Figure 17-18 shows this operation.

To ensure that no DC component is introduced on the panel, the SLEEP instruction should be executed immediately after a LCD frame boundary. The LCD interrupt can be used to determine the frame boundary. See **Section 17.9 "LCD Interrupts"** for the formulas to calculate the delay.

If a SLEEP instruction is executed and SLPEN = 0, the module will continue to display the current contents of the LCDDATA registers. To allow the module to continue operation while in Sleep, the clock source must be either the Timer1 oscillator or one of the

internal oscillators (either INTRC or INTOSC as the default system clock). While in Sleep, the LCD data cannot be changed. The LCD module current consumption will not decrease in this mode; however, the overall consumption of the device will be lower due to shut down of the core and other peripheral functions.

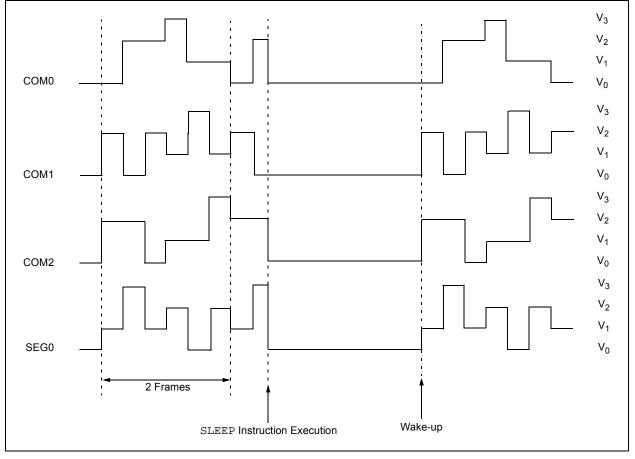
If the system clock is selected, and the module is not configured for Sleep operation, the module will ignore the SLPEN bit and stop operation immediately. The minimum LCD voltage will then be driven onto the segments and commons

17.10.1 USING THE LCD REGULATOR DURING SLEEP

Applications that use the LCD regulator for bias generation may not achieve the same degree of power reductions in Sleep mode when compared to applications using Mode 3 (resistor ladder) biasing. This is particularly true with Mode 0 operation, where the charge pump is active.

If Modes 0, 1 or 2 are used for bias generation, software contrast control will not be available.





17.11 Configuring the LCD Module

The following is the sequence of steps to configure the LCD module.

- 1. Select the frame clock prescale using bits, LP<3:0> (LCDPS<3:0>).
- 2. Configure the appropriate pins to function as segment drivers using the LCDSEx registers.
- 3. Configure the appropriate pins as inputs using the TRISx registers.
- 4. Configure the LCD module for the following using the LCDCON register:
 - Multiplex and Bias mode (LMUX<1:0>)
 - Timing source (CS<1:0>)
 - Sleep mode (SLPEN)
- 5. Write initial values to pixel data registers, LCDDATA0 through LCDDATA23.

- 6. Configure the LCD regulator:
 - a) If M2 or M3 bias configuration is to be used, turn off the regulator by setting CKSEL<1:0> (LCDREG<1:0>) to '00'. Set or clear the CPEN bit (LCDREG<6>) to select Mode 2 or Mode 3, as appropriate.
 - b) If M0 or M1 bias generation is to be used:
 - Set the VBIAS level using the BIAS<2:0> bits (LCDREG<5:3>).
 - Set or clear the CPEN bit to enable or disable the charge pump.
 - Set or clear the MODE13 bit (LCDREG<2>) to select the Bias mode.
 - Select a regulator clock source using the CKSEL<1:0> bits.
- 7. Clear LCD Interrupt Flag, LCDIF (PIR3<6>), and if desired, enable the interrupt by setting the LCDIE bit (PIE3<6>).
- 8. Enable the LCD module by setting the LCDEN bit (LCDCON<7>).

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59
PIR3	_	LCDIF	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF	62
PIE3	_	LCDIE	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE	62
IPR3	_	LCDIP	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP	62
RCON	IPEN	_	CM	RI	TO	PD	POR	BOR	60
LCDDATA23(1)	S47C3	S46C3	S45C3	S44C3	S43C3	S42C3	S41C3	S40C3	63
LCDDATA22	S39C3 ⁽¹⁾	S38C3 ⁽¹⁾	S37C3 ⁽¹⁾	S36C3 ⁽¹⁾	S35C3 ⁽¹⁾	S34C3 ⁽¹⁾	S33C3 ⁽¹⁾	S32C3	63
LCDDATA21	S31C3	S30C3	S29C3	S28C3	S27C3	S26C3	S25C3	S24C3	63
LCDDATA20	S23C3	S22C3	S21C3	S20C3	S19C3	S18C3	S17C3	S16C3	63
LCDDATA19	S15C3	S14C3	S13C3	S12C3	S11C3	S10C3	S09C3	S08C3	63
LCDDATA18	S07C3	S06C3	S05C3	S04C3	S03C3	S02C3	S01C3	S00C3	63
LCDDATA17 ⁽¹⁾	S47C2	S46C2	S45C2	S44C2	S43C2	S42C2	S41C2	S40C2	63
LCDDATA16	S39C2 ⁽¹⁾	S38C2 ⁽¹⁾	S37C2 ⁽¹⁾	S36C2 ⁽¹⁾	S35C2 ⁽¹⁾	S34C2 ⁽¹⁾	S33C2 ⁽¹⁾	S32C2	63
LCDDATA15	S31C2	S30C2	S29C2	S28C2	S27C2	S26C2	S25C2	S24C2	63
LCDDATA14	S23C2	S22C2	S21C2	S20C2	S19C2	S18C2	S17C2	S16C2	63
LCDDATA13	S15C2	S14C2	S13C2	S12C2	S11C2	S10C2	S09C2	S08C2	63
LCDDATA12	S07C2	S06C2	S05C2	S04C2	S03C2	S02C2	S01C2	S00C2	63
LCDDATA11 ⁽¹⁾	S47C1	S46C1	S45C1	S44C1	S43C1	S42C1	S41C1	S40C1	63
LCDDATA10	S39C1 ⁽¹⁾	S38C1 ⁽¹⁾	S37C1 ⁽¹⁾	S36C1 ⁽¹⁾	S35C1 ⁽¹⁾	S34C1 ⁽¹⁾	S33C1 ⁽¹⁾	S32C1	63
LCDDATA9	S31C1	S30C1	S29C1	S28C1	S27C1	S26C1	S25C1	S24C1	63
LCDDATA8	S23C1	S22C1	S21C1	S20C1	S19C1	S18C1	S17C1	S16C1	63
LCDDATA7	S15C1	S14C1	S13C1	S12C1	S11C1	S10C1	S09C1	S08C1	63
LCDDATA6	S07C1	S06C1	S05C1	S04C1	S03C1	S02C1	S01C1	S00C1	63
LCDDATA5 ⁽¹⁾	S47C0	S46C0	S45C0	S44C0	S43C0	S42C0	S41C0	S40C0	63
LCDDATA4	S39C0 ⁽¹⁾	S38C0 ⁽¹⁾	S37C0 ⁽¹⁾	S36C0 ⁽¹⁾	S35C0 ⁽¹⁾	S34C0 ⁽¹⁾	S33C0 ⁽¹⁾	S32C0	61
LCDDATA3	S31C0	S30C0	S29C0	S28C0	S27C0	S26C0	S25C0	S24C0	61
LCDDATA2	S23C0	S22C0	S21C0	S20C0	S19C0	S18C0	S17C0	S16C0	61
LCDDATA1	S15C0	S14C0	S13C0	S12C0	S11C0	S10C0	S09C0	S08C0	61
LCDDATA0	S07C0	S06C0	S05C0	S04C0	S03C0	S02C0	S01C0	S00C0	61
LCDSE5 ⁽¹⁾	SE47	SE46	SE45	SE44	SE43	SE42	SE41	SE40	61
LCDSE4	SE39 ⁽¹⁾	SE38 ⁽¹⁾	SE37 ⁽¹⁾	SE36 ⁽¹⁾	SE35 ⁽¹⁾	SE34 ⁽¹⁾	SE33 ⁽¹⁾	SE32	61
LCDSE3	SE31	SE30	SE29	SE28	SE27	SE26	SE25	SE24	61
LCDSE2	SE23	SE22	SE21	SE20	SE19	SE18	SE17	SE16	61
LCDSE1	SE15	SE14	SE13	SE12	SE11	SE10	SE09	SE08	61
LCDSE0	SE07	SE06	SE05	SE04	SE03	SE02	SE01	SE00	61
LCDCON	LCDEN	SLPEN	WERR		CS1	CS0	LMUX1	LMUX0	61
LCDPS	WFT	BIASMD	LCDA	WA	LP3	LP2	LP1	LP0	61
LCDREG	—	CPEN	BIAS2	BIAS1	BIAS0	MODE13	CKSEL1	CKSEL0	60

TABLE 17-6:	REGISTERS ASSOCIATED WITH LCD OPERATION
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Legend: — = unimplemented, read as '0'. Shaded cells are not used for LCD operation.

Note 1: These registers or individual bits are unimplemented on PIC18F6XJ90 devices.

NOTES:

18.0 MASTER SYNCHRONOUS SERIAL PORT (MSSP) MODULE

18.1 Master SSP (MSSP) Module Overview

The Master Synchronous Serial Port (MSSP) module is a serial interface, useful for communicating with other peripheral or microcontroller devices. These peripheral devices may be serial EEPROMs, shift registers, display drivers, A/D Converters, etc. The MSSP module can operate in one of two modes:

- Serial Peripheral Interface (SPI)
- Inter-Integrated Circuit (I²C[™])
 - Full Master mode
 - Slave mode (with general address call)

The I^2C interface supports the following modes in hardware:

- Master mode
- Multi-Master mode
- Slave mode

18.2 Control Registers

Each MSSP module has three associated control registers. These include a status register (SSPSTAT) and two control registers (SSPCON1 and SSPCON2). The use of these registers and their individual bits differ significantly depending on whether the MSSP module is operated in SPI or I^2C mode.

Additional details are provided under the individual sections.

18.3 SPI Mode

The SPI mode allows 8 bits of data to be synchronously transmitted and received simultaneously. All four modes of SPI are supported. To accomplish communication, typically three pins are used:

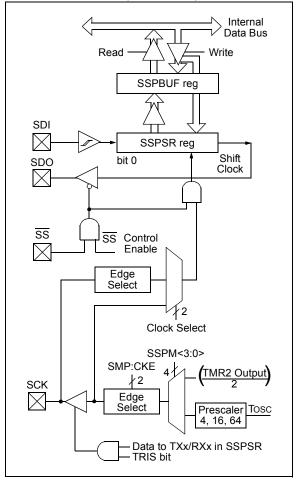
- Serial Data Out (SDO) RC5/SDO/SEG12
- Serial Data In (SDI) RC4/SDI/SDA/SEG16
- Serial Clock (SCK) RC3/SCK/SCL/SEG17

Additionally, a fourth pin may be used when in a Slave mode of operation:

• Slave Select (SS) – RF7/AN5/SS/SEG25

Figure 18-1 shows the block diagram of the MSSP module when operating in SPI mode.

FIGURE 18-1: MSSP BLOCK DIAGRAM (SPI MODE)



18.3.1 REGISTERS

Each MSSP module has four registers for SPI mode operation. These are:

- MSSP Control Register 1 (SSPCON1)
- MSSP Status Register (SSPSTAT)
- Serial Receive/Transmit Buffer Register (SSPBUF)
- MSSP Shift Register (SSPSR) Not directly accessible

SSPCON1 and SSPSTAT are the control and status registers in SPI mode operation. The SSPCON1 register is readable and writable. The lower 6 bits of the SSPSTAT are read-only. The upper two bits of the SSPSTAT are read/write.

SSPSR is the shift register used for shifting data in or out. SSPBUF is the buffer register to which data bytes are written to or read from.

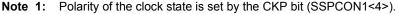
In receive operations, SSPSR and SSPBUF together, create a double-buffered receiver. When SSPSR receives a complete byte, it is transferred to SSPBUF and the SSPIF interrupt is set.

During transmission, the SSPBUF is not double-buffered. A write to SSPBUF will write to both, SSPBUF and SSPSR.

REGISTER 18-1: SSPSTAT: MSSP STATUS REGISTER (SPI MODE)

R/W-0	R/W-0	R-0	R-0	R-0	R-0	R0	R-0
SMP	CKE ⁽¹⁾	D/A	Р	S	R/W	UA	BF
bit 7							bit 0

Legend:							
R = Readable bit -n = Value at POR		W = Writable bit	U = Unimplemented bit, read as '0'				
		'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown			
bit 7	SMP: Sam	nple bit					
	SPI Maste	r mode:					
		data sampled at the end of o data sampled at the middle	•				
	<u>SPI Slave mode:</u> SMP must be cleared when SPI is used in Slave mode.						
bit 6	CKE: SPI Clock Select bit ⁽¹⁾						
	 1 = Transmit occurs on transition from active to Idle clock state 0 = Transmit occurs on transition from Idle to active clock state 						
bit 5	D/A: Data/Address bit						
	Used in I ²	C™ mode only.					
bit 4	P: Stop bit						
	Used in I ²	C mode only. This bit is clea	ared when the MSSP module is	s disabled; SSPEN is cleared.			
bit 3	S: Start bit	t					
	Used in I ²	C mode only.					
bit 2	R/W: Read/Write Information bit						
	Used in I ²	C mode only.					
bit 1		te Address bit					
	Used in I ²	C mode only.					
bit 0	BF: Buffer	Full Status bit (Receive mo	ode only)				
		ve complete; SSPBUF is ful					
	0 = Receiv	e not complete; SSPBUF is	s empty				



REGISTER 18-2: SSPCON1: MSSP CONTROL REGISTER 1 (SPI MODE)

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
WCOL	SSPOV ⁽¹⁾	SSPEN ⁽²⁾	CKP	SSPM3 ⁽³⁾	SSPM2 ⁽³⁾	SSPM1 ⁽³⁾	SSPM0 ⁽³⁾
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	WCOL: Write Collision Detect bit (Transmit mode only)
	 1 = The SSPBUF register is written while it is still transmitting the previous word (must be cleared in software)
	0 = No collision
bit 6	SSPOV: Receive Overflow Indicator bit ⁽¹⁾
	 SPI Slave mode: 1 = A new byte is received while the SSPBUF register is still holding the previous data. In case of overflow, the data in SSPSR is lost. Overflow can only occur in Slave mode. The user must read the SSPBUF, even if only transmitting data, to avoid setting overflow (must be cleared in software). 0 = No overflow
bit 5	SSPEN: Master Synchronous Serial Port Enable bit ⁽²⁾
	1 = Enables serial port and configures SCK, SDO, SDI and \overline{SS} as serial port pins 0 = Disables serial port and configures these pins as I/O port pins
bit 4	CKP: Clock Polarity Select bit
	1 = Idle state for clock is a high level
	0 = Idle state for clock is a low level
bit 3-0	SSPM<3:0>: Master Synchronous Serial Port Mode Select bits ⁽³⁾
	0101 = SPI Slave mode; clock = SCK pin, \overline{SS} pin control disabled, \overline{SS} can be used as I/O pin 0100 = SPI Slave mode; clock = SCK pin, \overline{SS} pin control enabled 0011 = SPI Master mode; clock = TMR2 output/2
	0010 = SPI Master mode; clock = Fosc/64
	0001 = SPI Master mode; clock = Fosc/16
	0000 = SPI Master mode; clock = Fosc/4
Note 1:	In Master mode, the overflow bit is not set since each new reception (and transmission) is initiated by writing to the SSPBUF register.

- 2: When enabled, these pins must be properly configured as inputs or outputs.
- **3:** Bit combinations not specifically listed here are either reserved or implemented in I^2C^{TM} mode only.

18.3.2 OPERATION

When initializing the SPI, several options need to be specified. This is done by programming the appropriate control bits (SSPCON1<5:0> and SSPSTAT<7:6>). These control bits allow the following to be specified:

- Master mode (SCK is the clock output)
- Slave mode (SCK is the clock input)
- Clock Polarity (Idle state of SCK)
- Data Input Sample Phase (middle or end of data output time)
- Clock Edge (output data on rising/falling edge of SCK)
- Clock Rate (Master mode only)
- Slave Select mode (Slave mode only)

Each MSSP consists of a transmit/receive shift register (SSPSR) and a buffer register (SSPBUF). The SSPSR shifts the data in and out of the device, MSb first. The SSPBUF holds the data that was written to the SSPSR until the received data is ready. Once the 8 bits of data have been received, that byte is moved to the SSPBUF register. Then, the Buffer Full detect bit, BF (SSPSTAT<0>), and the interrupt flag bit, SSPIF, are set. This double-buffering of the received data (SSPBUF) allows the next byte to start reception before

reading the data that was just received. Any write to the SSPBUF register during transmission/reception of data will be ignored and the Write Collision detect bit, WCOL (SSPCON1<7>), will be set. User software must clear the WCOL bit so that it can be determined if the following write(s) to the SSPBUF register completed successfully.

When the application software is expecting to receive valid data, the SSPBUF should be read before the next byte of data to transfer is written to the SSPBUF. The Buffer Full bit, BF (SSPSTAT<0>), indicates when SSPBUF has been loaded with the received data (transmission is complete). When the SSPBUF is read, the BF bit is cleared. This data may be irrelevant if the SPI is only a transmitter. Generally, the MSSP interrupt is used to determine when the transmission/reception has completed. The SSPBUF must be read and/or written. If the interrupt method is not going to be used, then software polling can be done to ensure that a write collision does not occur. Example 18-1 shows the loading of the SSPBUF (SSPSR) for data transmission.

The SSPSR is not directly readable or writable and can only be accessed by addressing the SSPBUF register. Additionally, the SSPSTAT register indicates the various status conditions.

EXAMPLE 18-1: LOADING THE SSPBUF (SSPSR) REGISTER

LOOP		SSPSTAT, BF	;Has data been received (transmit complete)?
	BRA	LOOP	;No
	MOVF	SSPBUF, W	;WREG reg = contents of SSPBUF
	MOVWF	RXDATA	;Save in user RAM, if data is meaningful
	MOVF	TXDATA, W	;W reg = contents of TXDATA
	MOVWF	SSPBUF	;New data to xmit

18.3.3 ENABLING SPI I/O

To enable the serial port, the MSSP Enable bit, SSPEN (SSPCON1<5>), must be set. To reset or reconfigure SPI mode, clear the SSPEN bit, reinitialize the SSPCON registers and then set the SSPEN bit. This configures the SDI, SDO, SCK and SS pins as serial port pins. For the pins to behave as the serial port function, some must have their data direction bits (in the TRIS register) appropriately programmed as follows:

- SDI is automatically controlled by the SPI module
- SDO must have TRISC<5> bit cleared
- SCK (Master mode) must have TRISC<3> bit cleared
- SCK (Slave mode) must have TRISC<3> bit set
- SS must have TRISF<7> bit set

Any serial port function that is not desired may be overridden by programming the corresponding data direction (TRIS) register to the opposite value.

18.3.4 OPEN-DRAIN OUTPUT OPTION

The drivers for the SDO output and SCK clock pins can be optionally configured as open-drain outputs. This feature allows the voltage level on the pin to be pulled to a higher level through an external pull-up resistor, and allows the output to communicate with external circuits without the need for additional level shifters.

The open-drain output option is controlled by the SPIOD bit (TRISG<7>). Setting this bit configures both pins for open-drain operation.

18.3.5 TYPICAL CONNECTION

Figure 18-2 shows a typical connection between two microcontrollers. The master controller (Processor 1) initiates the data transfer by sending the SCK signal. Data is shifted out of both shift registers on their programmed clock edge and latched on the opposite edge of the clock. Both processors should be programmed to the same Clock Polarity (CKP), then both controllers would send and receive data at the same time. Whether the data is meaningful (or dummy data) depends on the application software. This leads to three scenarios for data transmission:

- · Master sends data Slave sends dummy data
- · Master sends data Slave sends data
- Master sends dummy data Slave sends data

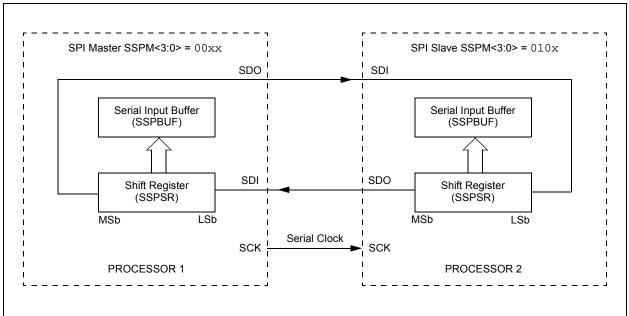


FIGURE 18-2: SPI MASTER/SLAVE CONNECTION

18.3.6 MASTER MODE

The master can initiate the data transfer at any time because it controls the SCK. The master determines when the slave (Processor 2, Figure 18-2) will broadcast data by the software protocol.

In Master mode, the data is transmitted/received as soon as the SSPBUF register is written to. If the SPI is only going to receive, the SDO output could be disabled (programmed as an input). The SSPSR register will continue to shift in the signal present on the SDI pin at the programmed clock rate. As each byte is received, it will be loaded into the SSPBUF register as if it was a normal received byte (interrupts and status bits appropriately set). This could be useful in receiver applications as a "Line Activity Monitor" mode. The clock polarity is selected by appropriately programming the CKP bit (SSPCON1<4>). This, then, would give waveforms for SPI communication, as shown in Figure 18-3, Figure 18-5 and Figure 18-6, where the MSB is transmitted first. In Master mode, the SPI clock rate (bit rate) is user-programmable to be one of the following:

- Fosc/4 (or Tcy)
- Fosc/16 (or 4 Tcy)
- Fosc/64 (or 16 Tcy)
- Timer2 output/2

This allows a maximum data rate (at 40 MHz) of 10.00 Mbps.

Figure 18-3 shows the waveforms for Master mode. When the CKE bit is set, the SDO data is valid before there is a clock edge on SCK. The change of the input sample is shown based on the state of the SMP bit. The time when the SSPBUF is loaded with the received data is shown.

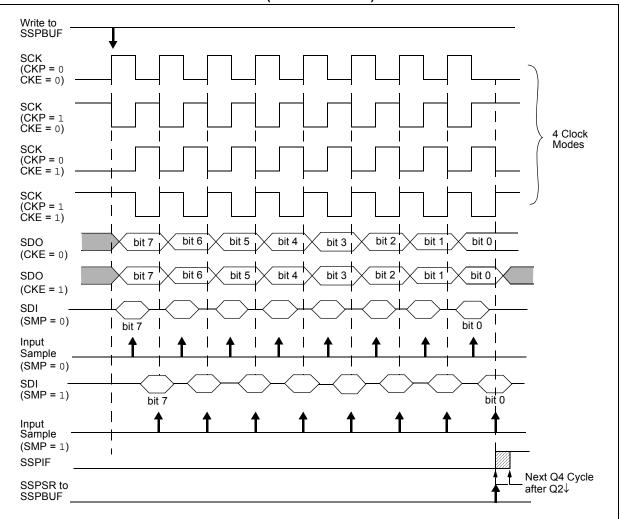


FIGURE 18-3: SPI MODE WAVEFORM (MASTER MODE)

18.3.7 SLAVE MODE

In Slave mode, the data is transmitted and received as the external clock pulses appear on SCK. When the last bit is latched, the SSPIF interrupt flag bit is set.

Before enabling the module in SPI Slave mode, the clock line must match the proper Idle state. The clock line can be observed by reading the SCK pin. The Idle state is determined by the CKP bit (SSPCON1<4>).

While in Slave mode, the external clock is supplied by the external clock source on the SCK pin. This external clock must meet the minimum high and low times as specified in the electrical specifications.

While in Sleep mode, the slave can transmit/receive data. When a byte is received, the device will wake-up from Sleep.

18.3.8 SLAVE SELECT SYNCHRONIZATION

The \overline{SS} pin allows a Synchronous Slave mode. The SPI must be in Slave mode with \overline{SS} pin control enabled (SSPCON1<3:0> = 04h). When the \overline{SS} pin is low, transmission and reception are enabled and the SDO

pin is driven. When the \overline{SS} pin goes high, the SDO pin is no longer driven, even if in the middle of a transmitted byte, and becomes a floating output. External pull-up/pull-down resistors may be desirable depending on the application.

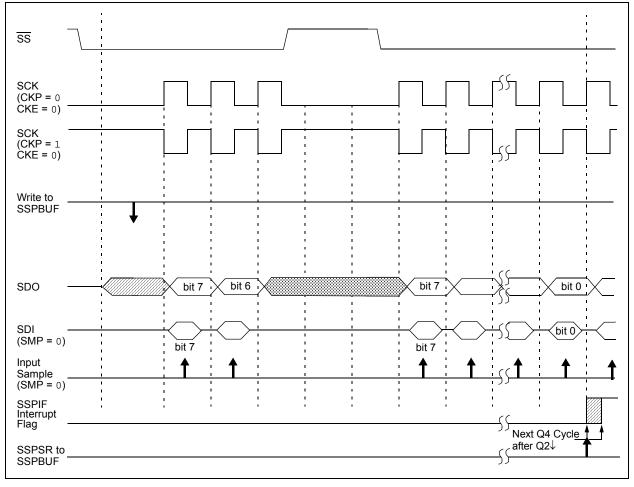
Note 1:	When	the	SPI	is	in	Slave	mode	with
	the	SS	pir	n	C	control	ena	abled
	(SSPC	ON1	<3:0	> =	01	00),	the	SPI
	module	e will	reset	if tl	ne s	SS pin i	s set to	VDD.
2.	If the S			1 in	912		do with	CKE

2: If the SPI is used in Slave mode with CKE set, then the SS pin control must be enabled.

When the SPI module resets, the bit counter is forced to '0'. This can be done by either forcing the \overline{SS} pin to a high level or clearing the SSPEN bit.

To emulate two-wire communication, the SDO pin can be connected to the SDI pin. When the SPI needs to operate as a receiver, the SDO pin can be configured as an input. This disables transmissions from the SDO. The SDI can always be left as an input (SDI function) since it cannot create a bus conflict.

FIGURE 18-4: SLAVE SYNCHRONIZATION WAVEFORM



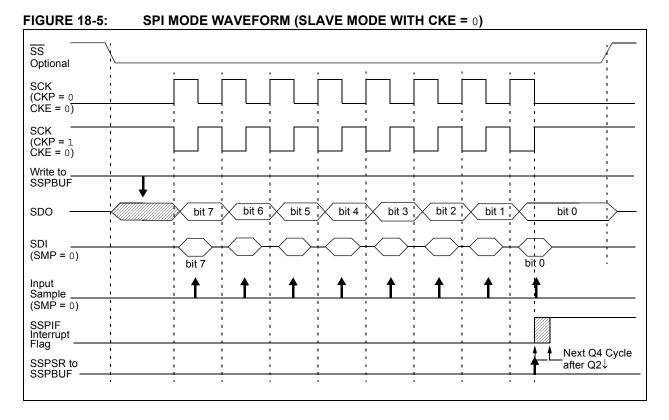
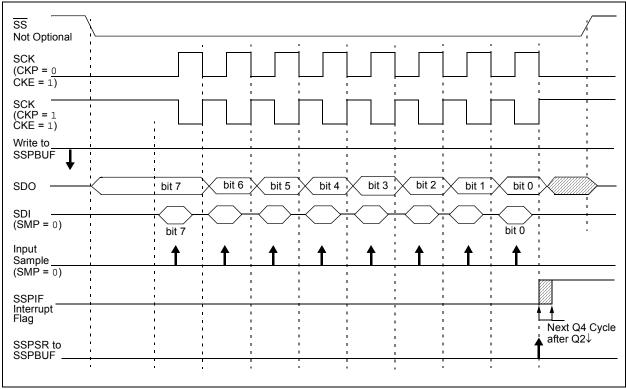


FIGURE 18-6: SPI MODE WAVEFORM (SLAVE MODE WITH CKE = 1)



18.3.9 OPERATION IN POWER-MANAGED MODES

In SPI Master mode, module clocks may be operating at a different speed than when in Full-Power mode; in the case of Sleep mode, all clocks are halted.

In Idle modes, a clock is provided to the peripherals. That clock should be from the primary clock source, the secondary clock (Timer1 oscillator at 32.768 kHz) or the INTRC source. See **Section 3.3 "Clock Sources and Oscillator Switching"** for additional information.

In most cases, the speed that the master clocks SPI data is not important; however, this should be evaluated for each system.

If MSSP interrupts are enabled, they can wake the controller from Sleep mode, or one of the Idle modes, when the master completes sending data. If an exit from Sleep or Idle mode is not desired, MSSP interrupts should be disabled.

If the Sleep mode is selected, all module clocks are halted and the transmission/reception will remain in that state until the device wakes. After the device returns to Run mode, the module will resume transmitting and receiving data.

In SPI Slave mode, the SPI Transmit/Receive Shift register operates asynchronously to the device. This allows the device to be placed in any power-managed mode and data to be shifted into the SPI Transmit/Receive Shift register. When all 8 bits have been received, the MSSP interrupt flag bit will be set, and if enabled, will wake the device.

18.3.10 EFFECTS OF A RESET

A Reset disables the MSSP module and terminates the current transfer.

18.3.11 BUS MODE COMPATIBILITY

Table 18-1 shows the compatibility between the standard SPI modes and the states of the CKP and CKE control bits.

TABLE 1	8-1:	SPI	BUS	MODES
		••••		

Standard SPI Mode	Control Bits State				
Terminology	СКР	CKE			
0, 0	0	1			
0, 1	0	0			
1, 0	1	1			
1, 1	1	0			

There is also an SMP bit which controls when the data is sampled.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59
PIR1	—	ADIF	RC1IF	TX1IF	SSPIF	—	TMR2IF	TMR1IF	62
PIE1	—	ADIE	RC1IE	TX1IE	SSPIE	_	TMR2IE	TMR1IE	62
IPR1	—	ADIP	RC1IP	TX1IP	SSPIP	—	TMR2IP	TMR1IP	62
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	62
TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1	—	62
TRISG	SPIOD	CCP2OD	CCP10D	TRISG4	TRISG3	TRISG2	TRISG1	TRISG0	62
SSPBUF MSSP Receive Buffer/Transmit Register								60	
SSPCON1	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	60
SSPSTAT	SMP	CKE	D/A	Р	S	R/W	UA	BF	60

TABLE 18-2: REGISTERS ASSOCIATED WITH SPI OPERATION

Legend: Shaded cells are not used by the MSSP module in SPI mode.

18.4 I²C Mode

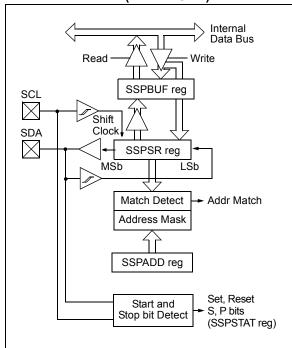
The MSSP module in I²C mode fully implements all master and slave functions (including general call support), and provides interrupts on Start and Stop bits in hardware to determine a free bus (multi-master function). The MSSP module implements the standard mode specifications as well as 7-bit and 10-bit addressing.

Two pins are used for data transfer:

- Serial clock (SCL) RC3/SCK/SCL
- Serial data (SDA) RC4/SDI/SDA

The user must configure these pins as inputs by setting the TRISC<4:3> bits.

FIGURE 18-7: MSSP BLOCK DIAGRAM (I²C™ MODE)



18.4.1 REGISTERS

The MSSP module has six registers for $\mathsf{I}^2\mathsf{C}$ operation. These are:

- MSSP Control Register 1 (SSPCON1)
- MSSP Control Register 2 (SSPCON2)
- MSSP Status Register (SSPSTAT)
- Serial Receive/Transmit Buffer Register (SSPBUF)
- MSSP Shift Register (SSPSR) Not directly accessible
- MSSP Address Register (SSPADD)

SSPCON1, SSPCON2 and SSPSTAT are the control and status registers in I^2C mode operation. The SSPCON1 and SSPCON2 registers are readable and writable. The lower 6 bits of the SSPSTAT are read-only. The upper two bits of the SSPSTAT are read/write.

Many of the bits in SSPCON2 assume different functions, depending on whether the module is operating in Master or Slave mode. The SSPCON2<5:2> bits also assume different names in Slave mode. The different aspects of SSPCON2 are shown in Register 18-5 (for Master mode) and Register 18-6 (Slave mode).

SSPSR is the shift register used for shifting data in or out. SSPBUF is the buffer register to which data bytes are written to or read from.

The SSPADD register holds the slave device address when the MSSP is configured in I^2C Slave mode. When the MSSP is configured in Master mode, the lower seven bits of SSPADD act as the Baud Rate Generator reload value.

In receive operations, SSPSR and SSPBUF together, create a double-buffered receiver. When SSPSR receives a complete byte, it is transferred to SSPBUF and the SSPIF interrupt is set.

During transmission, the SSPBUF is not double-buffered. A write to SSPBUF will write to both SSPBUF and SSPSR.

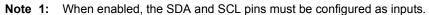
R/W-0	R/W-0	R-0	R-0	R-0	R-0	R-0	R-0				
SMP	CKE	D/A	P ⁽¹⁾	S ⁽¹⁾	R/W	UA	BF				
bit 7	ł			I			bit				
Lonondu											
Legend:	abla hit	\\/ = \\/ritabla	hit		nantad hit ra						
R = Reada		W = Writable '1' = Bit is set		U = Unimplen							
-n = Value				'0' = Bit is cle	areu	x = Bit is unkno	WII				
bit 7	SMP: Slew R	ate Control bit									
	In Master or S										
				d Speed mode (1 MHz)					
			a for High-Spe	eed mode (400 l	KHZ)						
bit 6	CKE: SMBus In Master or S										
		MBus specific	inputs								
		MBus specific									
bit 5	D/A: Data/Ad	dress bit									
	In Master mod	<u>de:</u>									
		Reserved.									
		In Slave mode: 1 = Indicates that the last byte received or transmitted was data									
 0 = Indicates that the last byte received or transmitted was address bit 4 P: Stop bit⁽¹⁾ 											
		1 = Indicates that a Stop bit has been detected last									
		0 = Stop bit was not detected last									
bit 3	S: Start bit ⁽¹⁾										
		that a Start bit		ected last							
		as not detecte									
bit 2		rite Informatio	n bit (l ² C™ mo	ode only)							
	<u>In Slave mode</u> 1 = Read	<u>ə:</u> (2)									
	1 = Read 0 = Write										
	In Master mod	de: ⁽³⁾									
	1 = Transmit i	is in progress									
		is not in progre									
bit 1	•	ddress bit (10		• /							
		that the user n loes not need		e the address in	the SSPADE) register					
bit 0	BF: Buffer Fu										
	<u>In Transmit m</u>										
	1 = SSPBUF										
	0 = SSPBUF										
	In Receive me 1 = SSPBUF		t include the $\frac{1}{4}$	ACK and Stop bi	ts)						
				ne ACK and Sto							
Note 1:	This bit is cleared	on Reset and	when SSPEN	is cleared.							
2:					s match. This	s bit is only valid fro	om the				
	address match to	the next Start	bit, Stop bit or	not ACK bit.							

2

3: ORing this bit with SEN, RSEN, PEN, RCEN or ACKEN will indicate if the MSSP is in Active mode.

	WCOL: Write In Master Tra 1 = A write t transmis 0 = No collis	to the SSPBUF sion to be starte	t bit	SSPM3 U = Unimpler '0' = Bit is cle	SSPM2 nented bit, rea ared	SSPM1 d as '0' x = Bit is unkr	SSPM0 bit 0			
Legend: R = Readable -n = Value at F bit 7	WCOL: Write In Master Tra 1 = A write t transmis 0 = No collis	'1' = Bit is set Collision Detections insmit mode: to the SSPBUF sion to be starte	t bit	-						
R = Readable -n = Value at F bit 7	WCOL: Write In Master Tra 1 = A write t transmis 0 = No collis	'1' = Bit is set Collision Detections insmit mode: to the SSPBUF sion to be starte	t bit	-			nown			
R = Readable -n = Value at F bit 7	WCOL: Write In Master Tra 1 = A write t transmis 0 = No collis	'1' = Bit is set Collision Detections insmit mode: to the SSPBUF sion to be starte	t bit	-			nown			
-n = Value at F bit 7	WCOL: Write In Master Tra 1 = A write t transmis 0 = No collis	'1' = Bit is set Collision Detections insmit mode: to the SSPBUF sion to be starte	t bit	-			nown			
bit 7	WCOL: Write In Master Tra 1 = A write t transmis 0 = No collis	e Collision Detect insmit mode: to the SSPBUF sion to be starte			ared	x = Bit is unkr	iown			
	In Master Tra 1 = A write t transmis 0 = No collis	insmit mode: to the SSPBUF sion to be starte								
	1 = A write t transmis 0 = No collis	to the SSPBUF sion to be starte	rogistor was							
	transmis 0 = No collis	sion to be starte	register wee							
	0 = No collis					onditions were	not valid for a			
		ION	a (must be ci	eared in soltwa	ie)					
		ismit mode:								
	 1 = The SSPBUF register is written while it is still transmitting the previous word (m software) 						t be cleared in			
		= No collision Receive mode (Master or Slave modes):								
	This is a "dor		Slave modes	<u>:</u>						
		eive Overflow In	dicator bit							
	In Receive mode: 1 = A byte is received while the SSPBUF register is still holding the pr software)					vious byte (mus	t be cleared in			
	0 = No overflow									
	In Transmit mode: This is a "don't care" bit in Transmit mode.									
	SSPEN: Master Synchronous Serial Port Enable bit ⁽¹⁾									
	 1 = Enables the serial port and configures the SDA and SCL pins as the serial port pins 0 = Disables serial port and configures these pins as I/O port pins 									
bit 4	CKP: SCK R	elease Control b	bit							
	In Slave mode:									
	 Release clock Holds clock low (clock stretch), used to ensure data setup time 									
	In Master mode:									
	Unused in thi	s mode.								
	$1111 = I^{2}C S$ $1110 = I^{2}C S$ $1011 = I^{2}C F$ $1000 = I^{2}C N$ $0111 = I^{2}C S$	Synchronous S lave mode, 10-b lave mode, 7-bit irmware Control faster mode, clo lave mode, 10-b	bit address wit t address with led Master m ck = Fosc/(4 bit address	th Start and Sto Start and Stop ode (slave Idle)	bit interrupts					
		lave mode, 7-bi		e are either rese	erved or implei	mented in SPI n	node only.			

REGISTER 18-4: SSPCON1: MSSP CONTROL REGISTER 1 (I²C[™] MODE)



R/W-0 R/W-0 R/W-0 R/W-0 R/W-0 R/W-0 R/W-0 R/W-0 ACKDT⁽¹⁾ ACKEN⁽²⁾ RCEN⁽²⁾ PEN⁽²⁾ RSEN⁽²⁾ SEN⁽²⁾ GCEN ACKSTAT bit 7 bit 0 Legend: W = Writable bit R = Readable bit U = Unimplemented bit, read as '0' -n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown bit 7 GCEN: General Call Enable bit Unused in Master mode. bit 6 ACKSTAT: Acknowledge Status bit (Master Transmit mode only) 1 = Acknowledge was not received from slave 0 = Acknowledge was received from slave ACKDT: Acknowledge Data bit (Master Receive mode only)⁽¹⁾ bit 5 1 = Not Acknowledge 0 = Acknowledge bit 4 ACKEN: Acknowledge Sequence Enable bit⁽²⁾ 1 = Initiate Acknowledge sequence on SDA and SCL pins and transmit ACKDT data bit; automatically cleared by hardware 0 = Acknowledge sequence Idle bit 3 RCEN: Receive Enable bit (Master Receive mode only)⁽²⁾ 1 = Enables Receive mode for I^2C^{TM} 0 = Receive Idle PEN: Stop Condition Enable bit⁽²⁾ bit 2 1 = Initiate Stop condition on SDA and SCL pins. Automatically cleared by hardware. 0 = Stop condition Idle RSEN: Repeated Start Condition Enable bit⁽²⁾ bit 1 1 = Initiate Repeated Start condition on SDA and SCL pins. Automatically cleared by hardware. 0 = Repeated Start condition Idle SEN: Start Condition Enable bit⁽²⁾ bit 0 1 = Initiate Start condition on SDA and SCL pins; automatically cleared by hardware 0 = Start condition Idle **Note 1:** Value that will be transmitted when the user initiates an Acknowledge sequence at the end of a receive.

REGISTER 18-5: SSPCON2: MSSP CONTROL REGISTER 2 (I²C™ MASTER MODE)

If the l²C module is active, these bits may not be set (no spooling) and the SSPBUF may not be written to (or writes to the SSPBUF are disabled).

			CONTROL						
R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0		
GCEN	ACKSTAT	ADMSK5	ADMSK4	ADMSK3	ADMSK2	ADMSK1	SEN ⁽¹⁾		
bit 7							bit (
Legend:									
R = Readab	le hit	W = Writable	bit	LI = LInimpler	nented bit, rea	1 as '0'			
-n = Value a		'1' = Bit is set		'0' = Bit is cle		x = Bit is unkn	own		
bit 7	GCEN: Gene	ral Call Enable	bit						
		terrupt when a all address dis	•	ldress (0000h)	is received in t	he SSPSR			
bit 6	ACKSTAT: Acknowledge Status bit								
	Unused in Slave mode.								
bit 5-2	ADMSK<5:2>: Slave Address Mask Select bits								
		of correspondir of correspondir							
bit 1	ADMSK1: Slave Address Least Significant bit(s) Mask Select bit								
	In 7-Bit Addre	essing mode:							
	0	of SSPADD<1>							
	0 = Masking of SSPADD<1> only is disabled								
	<u>In 10-Bit Addressing mode:</u> 1 = Masking of SSPADD<1:0> is enabled								
	0	of SSPADD<1:							
bit 0	SEN: Stretch								
			ed for both sla	ve transmit and	d slave receive	(stretch enable	d)		
	0 = Clock stre	etching is disab	led						
	_								

Note 1: If the I²C[™] module is active, this bit may not be set (no spooling) and the SSPBUF may not be written (or writes to the SSPBUF are disabled).

18.4.2 OPERATION

The MSSP module functions are enabled by setting the MSSP Enable bit, SSPEN (SSPCON1<5>).

The SSPCON1 register allows control of the I^2C operation. Four mode selection bits (SSPCON1<3:0>) allow one of the following I^2C modes to be selected:

- I²C Master mode, clock = (FOSC/4) x (SSPADD + 1)
- I²C Slave mode (7-bit address)
- I²C Slave mode (10-bit address)
- I²C Slave mode (7-bit address) with Start and Stop bit interrupts enabled
- I²C Slave mode (10-bit address) with Start and Stop bit interrupts enabled
- I²C Firmware Controlled Master mode, slave is Idle

Selection of any I²C mode, with the SSPEN bit set, forces the SCL and SDA pins to be open-drain, provided these pins are programmed to inputs by setting the appropriate TRISC or TRISD bits. To ensure proper operation of the module, pull-up resistors must be provided externally to the SCL and SDA pins.

18.4.3 SLAVE MODE

In Slave mode, the SCL and SDA pins must be configured as inputs (TRISC<4:3> set). The MSSP module will override the input state with the output data when required (slave-transmitter).

The I²C Slave mode hardware will always generate an interrupt on an exact address match. In addition, address masking will also allow the hardware to generate an interrupt for more than one address (up to 31 in 7-bit addressing and up to 63 in 10-bit addressing). Through the mode select bits, the user can also choose to interrupt on Start and Stop bits.

When an address is matched, or the data transfer after an address match is received, the hardware automatically will generate the Acknowledge (\overline{ACK}) pulse and load the SSPBUF register with the received value currently in the SSPSR register.

Any combination of the following conditions will cause the MSSP module not to give this ACK pulse:

- The Buffer Full bit, BF (SSPSTAT<0>), was set before the transfer was received.
- The overflow bit, SSPOV (SSPCON1<6>), was set before the transfer was received.

In this case, the SSPSR register value is not loaded into the SSPBUF, but bit, SSPIF, is set. The BF bit is cleared by reading the SSPBUF register, while bit, SSPOV, is cleared through software. The SCL clock input must have a minimum high and low for proper operation. The high and low times of the I^2C specification, as well as the requirement of the MSSP module, are shown in timing parameter 100 and parameter 101.

18.4.3.1 Addressing

Once the MSSP module has been enabled, it waits for a Start condition to occur. Following the Start condition, the 8 bits are shifted into the SSPSR register. All incoming bits are sampled with the rising edge of the clock (SCL) line. The value of register, SSPSR<7:1>, is compared to the value of the SSPADD register. The address is compared on the falling edge of the eighth clock (SCL) pulse. If the addresses match, and the BF and SSPOV bits are clear, the following events occur:

- 1. The SSPSR register value is loaded into the SSPBUF register.
- 2. The Buffer Full bit, BF, is set.
- 3. An ACK pulse is generated.
- 4. The MSSP Interrupt Flag bit, SSPIF, is set (and an interrupt is generated, if enabled) on the falling edge of the ninth SCL pulse.

In 10-Bit Addressing mode, two address bytes need to be received by the slave. The five Most Significant bits (MSbs) of the first address byte specify if this is a 10-bit address. The R/W (SSPSTAT<2>) bit must specify a write so the slave device will receive the second address byte. For a 10-bit address, the first byte would equal '11110 A9 A8 0', where 'A9' and 'A8' are the two MSbs of the address. The sequence of events for 10-bit addressing is as follows, with steps 7 through 9 for the slave-transmitter:

- 1. Receive first (high) byte of address (bits, SSPIF, BF and UA (SSPSTAT<1>), are set).
- 2. Update the SSPADD register with second (low) byte of address (clears bit, UA, and releases the SCL line).
- 3. Read the SSPBUF register (clears bit, BF) and clear flag bit, SSPIF.
- 4. Receive second (low) byte of address (SSPIF, BF and UA bits are set).
- 5. Update the SSPADD register with the first (high) byte of address. If match releases SCL line, this will clear the UA bit.
- 6. Read the SSPBUF register (clears bit, BF) and clear flag bit, SSPIF.
- 7. Receive Repeated Start condition.
- 8. Receive first (high) byte of address (SSPIF and BF bits are set).
- 9. Read the SSPBUF register (clears BF bit) and clear flag bit, SSPIF.

18.4.3.2 Address Masking

Masking an address bit causes that bit to become a "don't care". When one address bit is masked, two addresses will be Acknowledged and cause an interrupt. It is possible to mask more than one address bit at a time, which makes it possible to Acknowledge up to 31 addresses in 7-bit mode and up to 63 addresses in 10-bit mode (see Example 18-2).

The l^2C slave behaves the same way, whether address masking is used or not. However, when address masking is used, the l^2C slave can Acknowledge multiple addresses and cause interrupts. When this occurs, it is necessary to determine which address caused the interrupt by checking SSPBUF.

In 7-Bit Addressing mode, address mask bits, ADMSK<5:1> (SSPCON<5:1>), mask the corresponding address bits in the SSPADD register. For any ADMSK bits that are set (ADMSK<n> = 1), the corresponding address bit is ignored (SSPADD<n> = x). For the module to issue an address Acknowledge, it is sufficient to match only on addresses that do not have an active address mask.

EXAMPLE 18-2: ADDRESS MASKING EXAMPLES

7-Bit Addressing:

SSPADD<7:1> = A0h (1010000) (SSPADD<0> is assumed to be '0')

ADMSK<5:1> = 00111

Addresses Acknowledged: A0h, A2h, A4h, A6h, A8h, AAh, ACh, AEh

10-Bit Addressing:

SSPADD<7:0> = A0h (10100000) (the two MSbs of the address are ignored in this example, since they are not affected by masking)

ADMSK<5:1> = 00111

Addresses Acknowledged: A0h, A1h, A2h, A3h, A4h, A5h, A6h, A7h, A8h, A9h, AAh, ABh, ACh, ADh, AEh, AFh

In 10-Bit Addressing mode, the ADMSK<5:2> bits mask the corresponding address bits in the SSPADD register. In addition, ADMSK1 simultaneously masks the two LSbs of the address (SSPADD<1:0>). For any ADMSK bits that are active (ADMSK<n> = 1), the corresponding address bit is ignored (SSPADD<n> = x). Also note, that although in 10-Bit Addressing mode, the upper address bits reuse part of the SSPADD register bits. The address mask bits do not interact with those bits. They only affect the lower address bits.

Note 1: ADMSK1 masks the two Least Significant bits of the address.

 The two Most Significant bits of the address are not affected by address masking.

18.4.3.3 Reception

When the R/\overline{W} bit of the address byte is clear and an address match occurs, the R/\overline{W} bit of the SSPSTAT register is cleared. The received address is loaded into the SSPBUF register and the SDA line is held low (ACK).

When the address byte overflow condition exists, then the no Acknowledge (ACK) pulse is given. An overflow condition is defined as either bit, BF (SSPSTAT<0>), is set or bit, SSPOV (SSPCON1<6>), is set.

An MSSP interrupt is generated for each data transfer byte. The interrupt flag bit, SSPIF, must be cleared in software. The SSPSTAT register is used to determine the status of the byte.

If SEN is enabled (SSPCON2<0> = 1), SCK/SCL will be held low (clock stretch) following each data transfer. The clock must be released by setting bit, CKP (SSPCON1<4>). See **Section 18.4.4 "Clock Stretching"** for more details.

18.4.3.4 Transmission

When the R/\overline{W} bit of the incoming address byte is set and an address match occurs, the R/\overline{W} bit of the SSPSTAT register is set. The received address is loaded into the SSPBUF register. The ACK pulse will be sent on the ninth bit and pin, RC3, is held low, regardless of SEN (see **Section 18.4.4 "Clock Stretching"** for more details). By stretching the clock, the master will be unable to assert another clock pulse until the slave is done preparing the transmit data. The transmit data must be loaded into the SSPBUF register which also loads the SSPSR register. Then, pin, RC3, should be enabled by setting bit, CKP (SSPCON1<4>). The eight data bits are shifted out on the falling edge of the SCL input. This ensures that the SDA signal is valid during the SCL high time (Figure 18-10).

The ACK pulse from the master-receiver is latched on the rising edge of the ninth SCL input pulse. If the SDA line is high (not ACK), then the data transfer is complete. In this case, when the ACK is latched by the slave, the slave logic is reset and the slave monitors for another occurrence of the Start bit. If the SDA line was low (ACK), the next transmit data must be loaded into the SSPBUF register. Again, pin, RC3, must be enabled by setting bit, CKP.

An MSSP interrupt is generated for each data transfer byte. The SSPIF bit must be cleared in software and the SSPSTAT register is used to determine the status of the byte. The SSPIF bit is set on the falling edge of the ninth clock pulse.

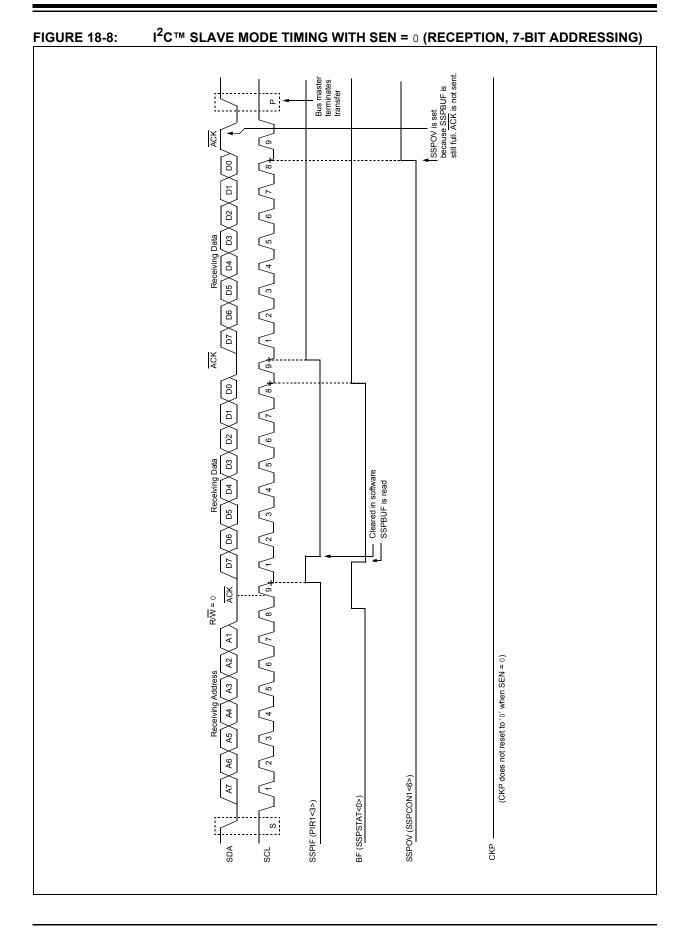
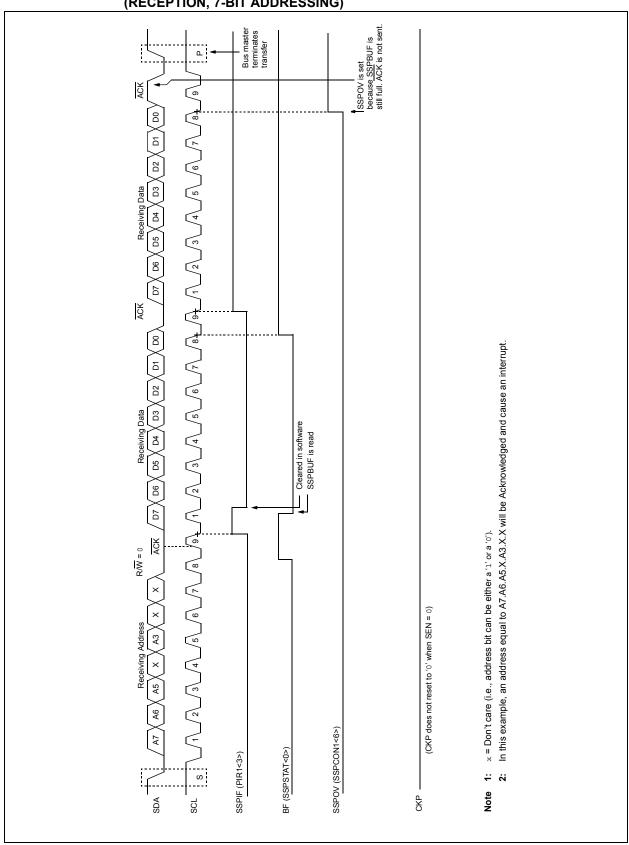
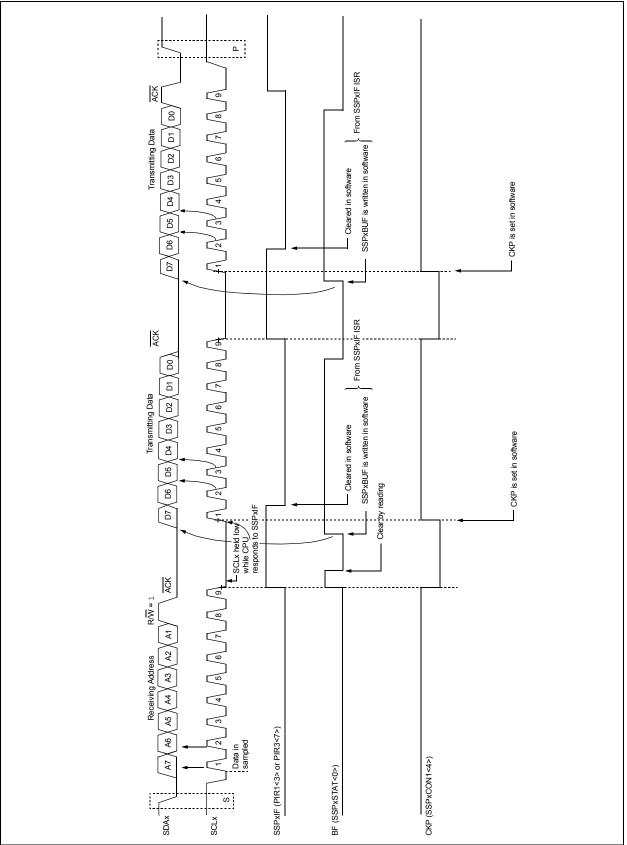
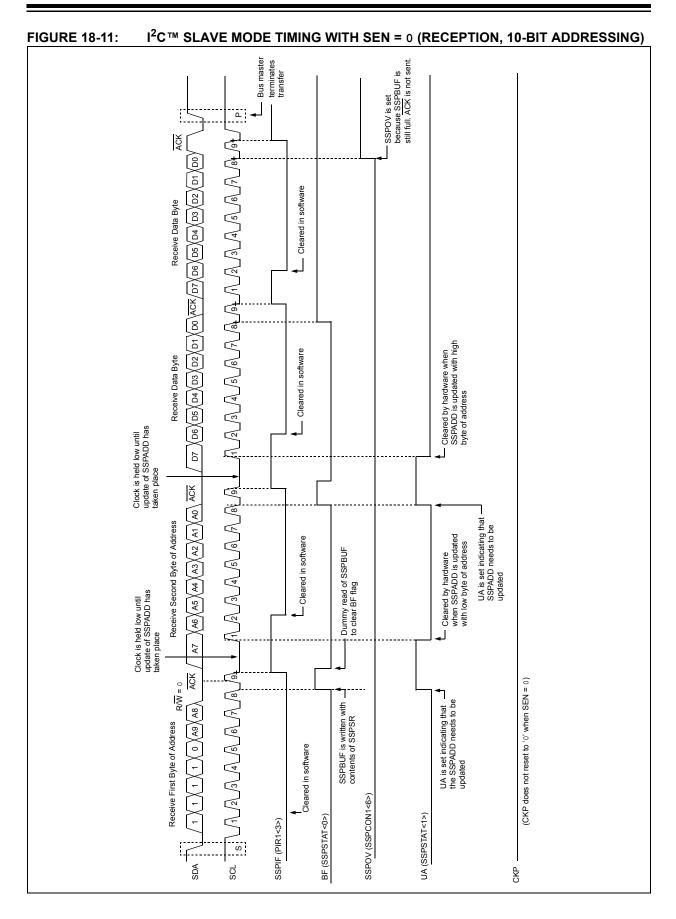


FIGURE 18-9: I²C[™] SLAVE MODE TIMING WITH SEN = 0 AND ADMSK<5:1> = 01011 (RECEPTION, 7-BIT ADDRESSING)









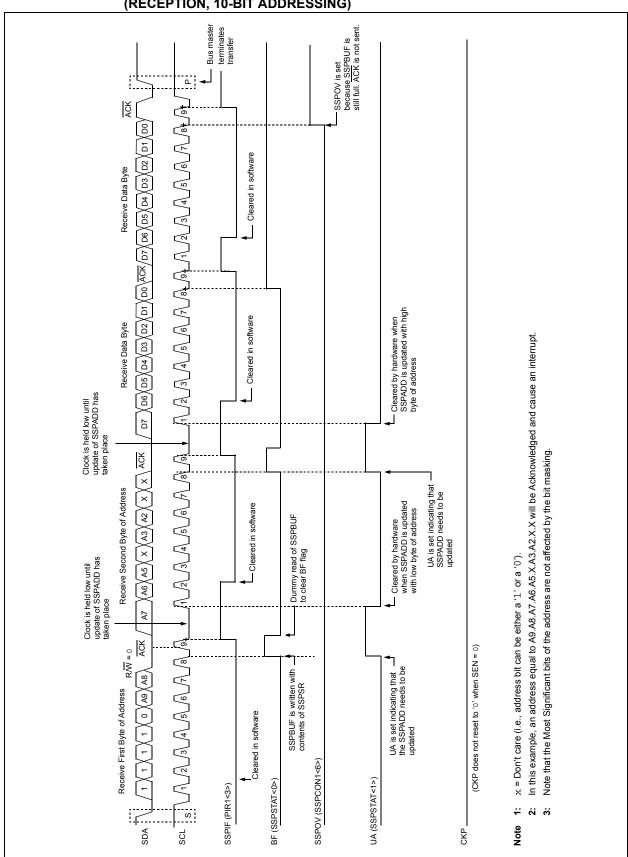
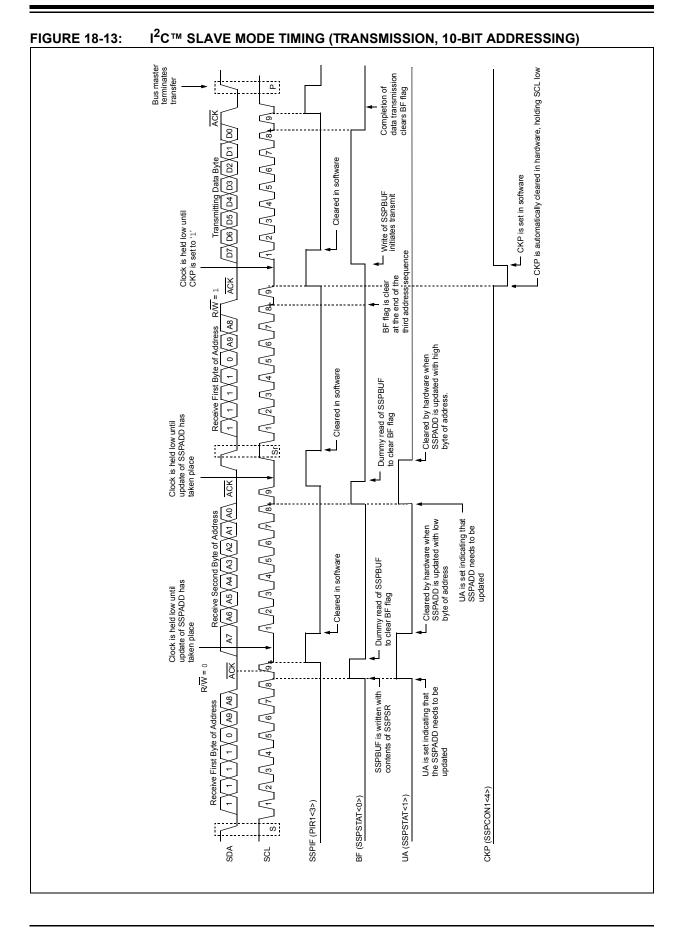


FIGURE 18-12: I²C[™] SLAVE MODE TIMING WITH SEN = 0 AND ADMSK<5:1> = 01001 (RECEPTION, 10-BIT ADDRESSING)



18.4.4 CLOCK STRETCHING

Both 7-Bit and 10-Bit Slave modes implement automatic clock stretching during a transmit sequence.

The SEN bit (SSPCON2<0>) allows clock stretching to be enabled during receives. Setting SEN will cause the SCL pin to be held low at the end of each data receive sequence.

18.4.4.1 Clock Stretching for 7-Bit Slave Receive Mode (SEN = 1)

In 7-Bit Slave Receive mode, on the falling edge of the ninth clock at the end of the ACK sequence, if the BF bit is set, the CKP bit in the SSPCON1 register is automatically cleared, forcing the SCL output to be held low. The CKP being cleared to '0' will assert the SCL line low. The CKP bit must be set in the user's ISR before reception is allowed to continue. By holding the SCL line low, the user has time to service the ISR and read the contents of the SSPBUF before the master device can initiate another receive sequence. This will prevent buffer overruns from occurring (see Figure 18-15).

- Note 1: If the user reads the contents of the SSPBUF before the falling edge of the ninth clock, thus clearing the BF bit, the CKP bit will not be cleared and clock stretching will not occur.
 - 2: The CKP bit can be set in software regardless of the state of the BF bit. The user should be careful to clear the BF bit in the ISR before the next receive sequence in order to prevent an overflow condition.

18.4.4.2 Clock Stretching for 10-Bit Slave Receive Mode (SEN = 1)

In 10-Bit Slave Receive mode, during the address sequence, clock stretching automatically takes place but CKP is not cleared. During this time, if the UA bit is set after the ninth clock, clock stretching is initiated. The UA bit is set after receiving the upper byte of the 10-bit address, and following the receive of the second byte of the 10-bit address, with the R/W bit cleared to '0'. The release of the clock line occurs upon updating SSPADD. Clock stretching will occur on each data receive sequence as described in 7-bit mode.

Note: If the user polls the UA bit and clears it by updating the SSPADD register before the falling edge of the ninth clock occurs, and if the user hasn't cleared the BF bit by reading the SSPBUF register before that time, then the CKP bit will still NOT be asserted low. Clock stretching on the basis of the state of the BF bit only occurs during a data sequence, not an address sequence.

18.4.4.3 Clock Stretching for 7-Bit Slave Transmit Mode

The 7-Bit Slave Transmit mode implements clock stretching by clearing the CKP bit after the falling edge of the ninth clock if the BF bit is clear. This occurs regardless of the state of the SEN bit.

The user's ISR must set the CKP bit before transmission is allowed to continue. By holding the SCL line low, the user has time to service the ISR and load the contents of the SSPBUF before the master device can initiate another transmit sequence (see Figure 18-10).

- Note 1: If the user loads the contents of SSPBUF, setting the BF bit before the falling edge of the ninth clock, the CKP bit will not be cleared and clock stretching will not occur.
 - 2: The CKP bit can be set in software regardless of the state of the BF bit.

18.4.4.4 Clock Stretching for 10-Bit Slave Transmit Mode

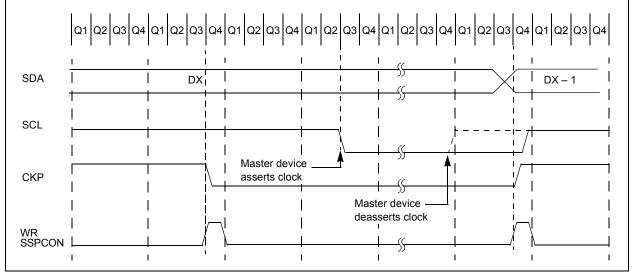
In 10-Bit Slave Transmit mode, clock stretching is controlled during the first two address sequences by the state of the UA bit, just as it is in 10-Bit Slave Receive mode. The first two addresses are followed by a third address sequence which contains the high-order bits of the 10-bit address and the R/W bit set to '1'. After the third address sequence is performed, the UA bit is not set. The module is now configured in Transmit mode and clock stretching is controlled by the BF flag as in 7-Bit Slave Transmit mode (see Figure 18-13).

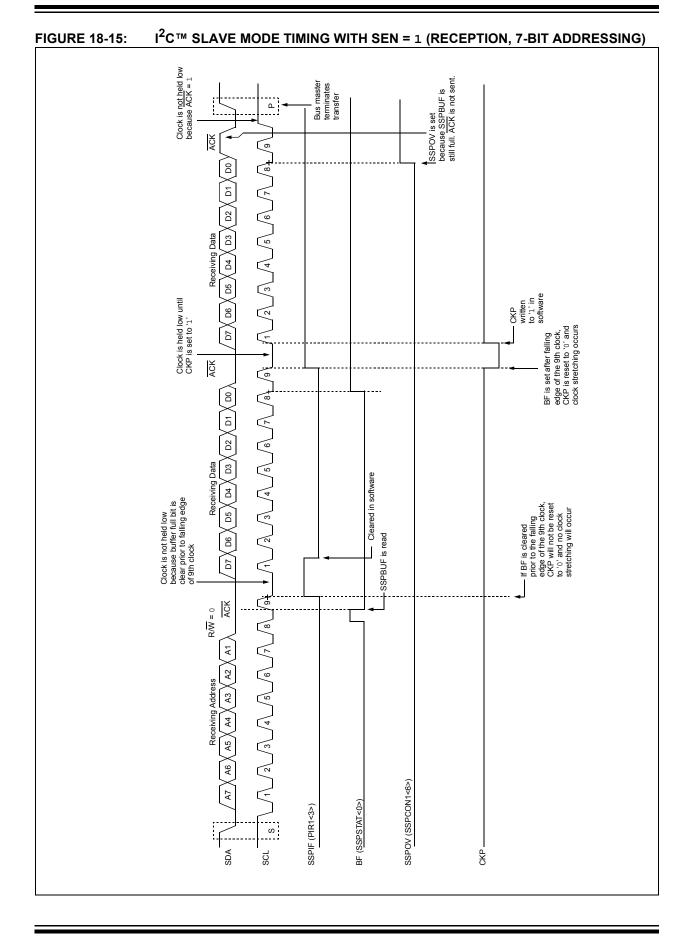
18.4.4.5 Clock Synchronization and the CKP bit

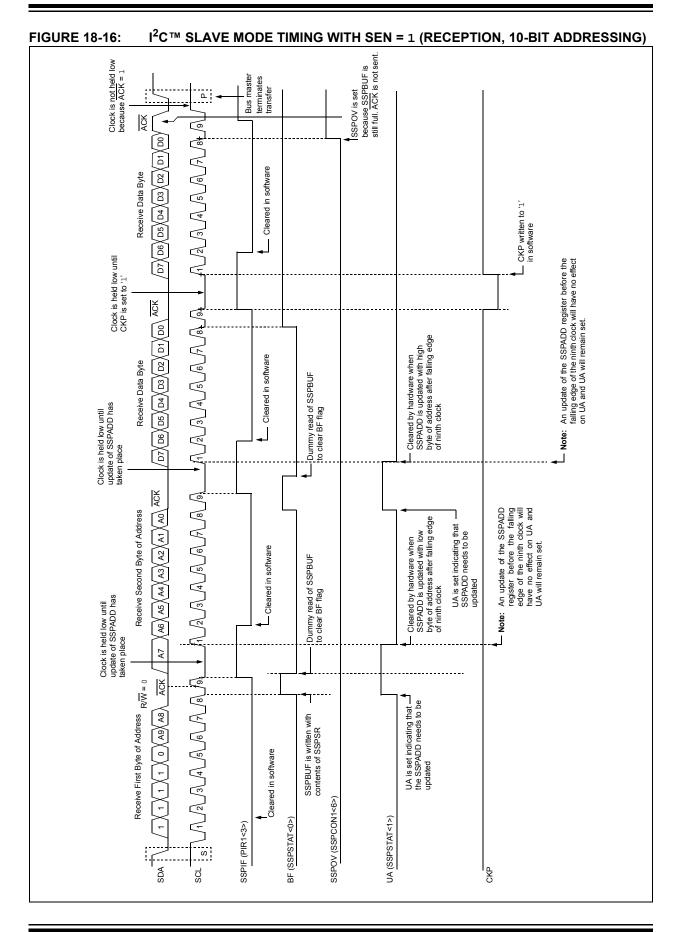
When the CKP bit is cleared, the SCL output is forced to '0'. However, clearing the CKP bit will not assert the SCL output low until the SCL output is already sampled low. Therefore, the CKP bit will not assert the SCL line until an external I^2C master device has

already asserted the SCL line. The SCL output will remain low until the CKP bit is set and all other devices on the I^2 C bus have deasserted SCL. This ensures that a write to the CKP bit will not violate the minimum high time requirement for SCL (see Figure 18-14).









18.4.5 GENERAL CALL ADDRESS SUPPORT

The addressing procedure for the I²C bus is such that the first byte after the Start condition usually determines which device will be the slave addressed by the master. The exception is the general call address which can address all devices. When this address is used, all devices should, in theory, respond with an Acknowledge.

The general call address is one of eight addresses reserved for specific purposes by the I^2C protocol. It consists of all '0's with R/W = 0.

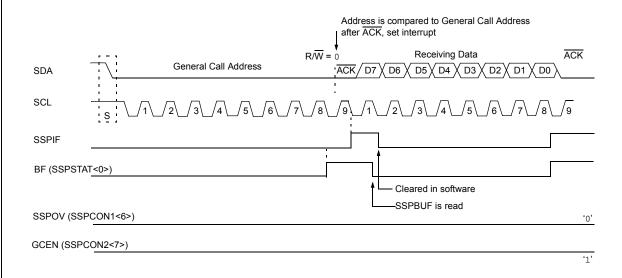
The general call address is recognized when the General Call Enable bit, GCEN, is enabled (SSPCON2<7> set). Following a Start bit detect, 8 bits are shifted into the SSPSR and the address is compared against the SSPADD. It is also compared to the general call address and fixed in hardware.

If the general call address matches, the SSPSR is transferred to the SSPBUF, the BF flag bit is set (eighth bit), and on the falling edge of the ninth bit (ACK bit), the SSPIF interrupt flag bit is set.

When the interrupt is serviced, the source for the interrupt can be checked by reading the contents of the SSPBUF. The value can be used to determine if the address was device-specific or a general call address.

In 10-bit mode, the SSPADD is required to be updated for the second half of the address to match and the UA bit is set (SSPSTAT<1>). If the general call address is sampled when the GCEN bit is set, while the slave is configured in 10-Bit Addressing mode, then the second half of the address is not necessary, the UA bit will not be set and the slave will begin receiving data after the Acknowledge (Figure 18-17).





18.4.6 MASTER MODE

Master mode is enabled by setting and clearing the appropriate SSPM bits in SSPCON1 and by setting the SSPEN bit. In Master mode, the SCL and SDA lines are manipulated by the MSSP hardware.

Master mode of operation is supported by interrupt generation on the detection of the Start and Stop conditions. The Stop (P) and Start (S) bits are cleared from a Reset or when the MSSP module is disabled. Control of the I^2C bus may be taken when the P bit is set, or the bus is Idle, with both the S and P bits clear.

In Firmware Controlled Master mode, user code conducts all I^2C bus operations based on Start and Stop bit conditions.

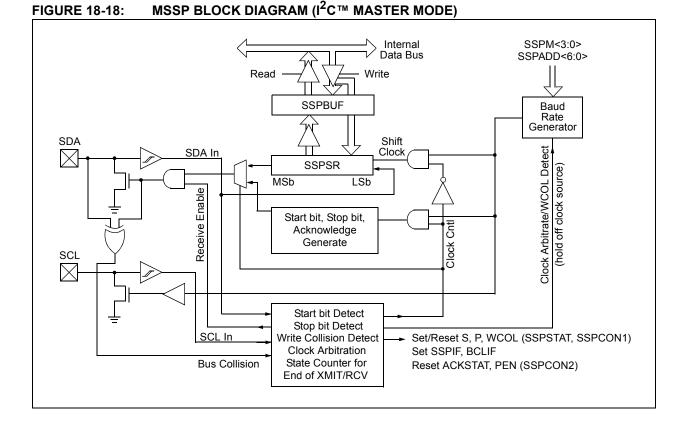
Once Master mode is enabled, the user has six options.

- 1. Assert a Start condition on SDA and SCL.
- 2. Assert a Repeated Start condition on SDA and SCL.
- 3. Write to the SSPBUF register initiating transmission of data/address.
- 4. Configure the I²C port to receive data.
- 5. Generate an Acknowledge condition at the end of a received byte of data.
- 6. Generate a Stop condition on SDA and SCL.

Note: The MSSP module, when configured in I²C Master mode, does not allow queueing of events. For instance, the user is not allowed to initiate a Start condition and immediately write the SSPBUF register to initiate transmission before the Start condition is complete. In this case, the SSPBUF will not be written to and the WCOL bit will be set, indicating that a write to the SSPBUF did not occur.

The following events will cause the MSSP Interrupt Flag bit, SSPIF, to be set (and an MSSP interrupt, if enabled):

- Start condition
- Stop condition
- · Data transfer byte transmitted/received
- · Acknowledge transmit
- Repeated Start



18.4.6.1 I²C Master Mode Operation

The master device generates all of the serial clock pulses and the Start and Stop conditions. A transfer is ended with a Stop condition or with a Repeated Start condition. Since the Repeated Start condition is also the beginning of the next serial transfer, the I²C bus will not be released.

In Master Transmitter mode, serial data is output through SDA, while SCL outputs the serial clock. The first byte transmitted contains the slave address of the receiving device (7 bits) and the Read/Write (R/W) bit. In this case, the R/W bit will be logic '0'. Serial data is transmitted, 8 bits at a time. After each byte is transmitted, an Acknowledge bit is received. Start and Stop conditions are output to indicate the beginning and the end of a serial transfer.

In Master Receive mode, the first byte transmitted contains the slave address of the transmitting device (7 bits) and the R/\overline{W} bit. In this case, the R/\overline{W} bit will be logic '1'. Thus, the first byte transmitted is a 7-bit slave address followed by a '1' to indicate the receive bit. Serial data is received via SDA, while SCL outputs the serial clock. Serial data is received, 8 bits at a time. After each byte is received, an Acknowledge bit is transmitted. Start and Stop conditions indicate the beginning and end of transmission.

The Baud Rate Generator used for the SPI mode operation is used to set the SCL clock frequency for either 100 kHz, 400 kHz or 1 MHz I²C operation. See **Section 18.4.7 "Baud Rate"** for more detail.

A typical transmit sequence would go as follows:

- 1. The user generates a Start condition by setting the Start Enable bit, SEN (SSPCON2<0>).
- 2. SSPIF is set. The MSSP module will wait the required start time before any other operation takes place.
- 3. The user loads the SSPBUF with the slave address to transmit.
- 4. The address is shifted out on the SDA pin until all 8 bits are transmitted.
- 5. The MSSP module shifts in the ACK bit from the slave device and writes its value into the SSPCON2 register (SSPCON2<6>).
- 6. The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSPIF bit.
- 7. The user loads the SSPBUF with eight bits of data.
- 8. Data is shifted out on the SDA pin until all 8 bits are transmitted.
- The MSSP module shifts in the ACK bit from the slave device and writes its value into the SSPCON2 register (SSPCON2<6>).
- 10. The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSPIF bit.
- 11. The user generates a Stop condition by setting the Stop Enable bit, PEN (SSPCON2<2>).
- 12. An interrupt is generated once the Stop condition is complete.

18.4.7 BAUD RATE

In I²C Master mode, the Baud Rate Generator (BRG) reload value is placed in the lower 7 bits of the SSPADD register (Figure 18-19). When a write occurs to SSPBUF, the Baud Rate Generator will automatically begin counting. The BRG counts down to 0 and stops until another reload has taken place. The BRG count is decremented twice per instruction cycle (TcY) on the Q2 and Q4 clocks. In I²C Master mode, the BRG is reloaded automatically.

Once the given operation is complete (i.e., transmission of the last data bit is followed by ACK), the internal clock will automatically stop counting and the SCL pin will remain in its last state.

Table 18-3 demonstrates clock rates based on instruction cycles and the BRG value loaded into SSPADD.

18.4.7.1 Baud Rate Generation in Power-Managed Modes

When the device is operating in one of the power-managed modes, the clock source to the BRG may change frequency, or even stop, depending on the mode and clock source selected. Switching to a Run or Idle mode from either the secondary clock or internal oscillator is likely to change the clock rate to the BRG. In Sleep mode, the BRG will not be clocked at all.

FIGURE 18-19: BAUD RATE GENERATOR BLOCK DIAGRAM

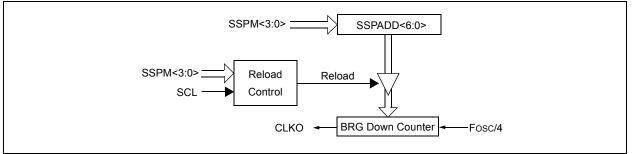


TABLE 18-3: I²C[™] CLOCK RATE w/BRG

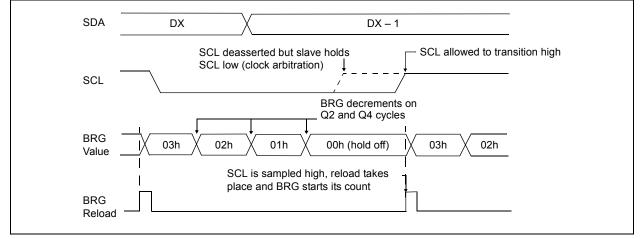
Fcy	Fcy * 2	BRG Value	FscL (2 Rollovers of BRG)
10 MHz	20 MHz	18h	400 kHz
10 MHz	20 MHz	1Fh	312.5 kHz
10 MHz	20 MHz	63h	100 kHz
4 MHz	8 MHz	09h	400 kHz
4 MHz	8 MHz	0Ch	308 kHz
4 MHz	8 MHz	27h	100 kHz
1 MHz	2 MHz	02h	333 kHz
1 MHz	2 MHz	09h	100 kHz
1 MHz	2 MHz	00h	1 MHz

18.4.7.2 Clock Arbitration

Clock arbitration occurs when the master, during any receive, transmit or Repeated Start/Stop condition, deasserts the SCL pin (SCL allowed to float high). When the SCL pin is allowed to float high, the Baud Rate Generator (BRG) is suspended from counting until the SCL pin is actually sampled high. When the

SCL pin is sampled high, the Baud Rate Generator is reloaded with the contents of SSPADD<6:0> and begins counting. This ensures that the SCL high time will always be at least one BRG rollover count in the event that the clock is held low by an external device (Figure 18-20).





18.4.8 I²C MASTER MODE START CONDITION TIMING

To initiate a Start condition, the user sets the Start Enable bit, SEN (SSPCON2<0>). If the SDA and SCL pins are sampled high, the Baud Rate Generator is reloaded with the contents of SSPADD<6:0> and starts its count. If SCL and SDA are both sampled high when the Baud Rate Generator times out (TBRG), the SDA pin is driven low. The action of the SDA pin being driven low while SCL is high is the Start condition and causes the S bit (SSPSTAT<3>) to be set. Following this, the Baud Rate Generator is reloaded with the contents of SSPADD<6:0> and resumes its count. When the Baud Rate Generator times out (TBRG), the SEN bit (SSPCON2<0>) will be automatically cleared by hardware. The Baud Rate Generator is suspended, leaving the SDA line held low and the Start condition is complete.

Note: If, at the beginning of the Start condition, the SDA and SCL pins are already sampled low, or if during the Start condition, the SCL line is sampled low before the SDA line is driven low, a bus collision occurs. The Bus Collision Interrupt Flag, BCLIF, is set, the Start condition is aborted and the I²C module is reset into its Idle state.

18.4.8.1 WCOL Status Flag

If the user writes the SSPBUF when a Start sequence is in progress, the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur).

Note: Because queueing of events is not allowed, writing to the lower 5 bits of SSPCON2 is disabled until the Start condition is complete.

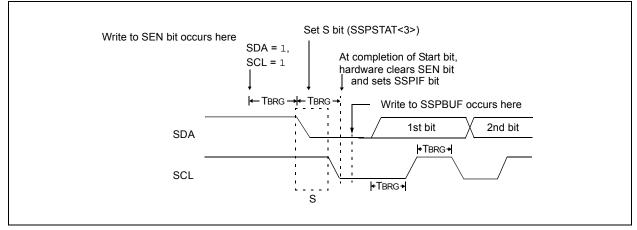


FIGURE 18-21: FIRST START BIT TIMING

18.4.9 I²C MASTER MODE REPEATED START CONDITION TIMING

A Repeated Start condition occurs when the RSEN bit (SSPCON2<1>) is programmed high and the I²C logic module is in the Idle state. When the RSEN bit is set, the SCL pin is asserted low. When the SCL pin is sampled low, the Baud Rate Generator is loaded with the contents of SSPADD<6:0> and begins counting. The SDA pin is released (brought high) for one Baud Rate Generator count (TBRG). When the Baud Rate Generator times out, if SDA is sampled high, the SCL pin will be deasserted (brought high). When SCL is sampled high, the Baud Rate Generator is reloaded with the contents of SSPADD<6:0> and begins counting. SDA and SCL must be sampled high for one TBRG. This action is then followed by assertion of the SDA pin (SDA = 0) for one TBRG while SCL is high. Following this, the RSEN bit (SSPCON2<1>) will be automatically cleared and the Baud Rate Generator will not be reloaded, leaving the SDA pin held low. As soon as a Start condition is detected on the SDA and SCL pins, the S bit (SSPSTAT<3>) will be set. The SSPIF bit will not be set until the Baud Rate Generator has timed out.

- **Note 1:** If RSEN is programmed while any other event is in progress, it will not take effect.
 - **2:** A bus collision during the Repeated Start condition occurs if:
 - SDA is sampled low when SCL goes from low-to-high.
 - SCL goes low before SDA is asserted low. This may indicate that another master is attempting to transmit a data '1'.

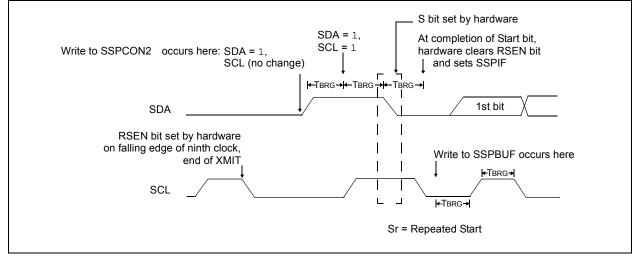
Immediately following the SSPIF bit getting set, the user may write the SSPBUF with the 7-bit address in 7-bit mode or the default first address in 10-bit mode. After the first eight bits are transmitted and an ACK is received, the user may then transmit an additional eight bits of address (10-bit mode) or eight bits of data (7-bit mode).

18.4.9.1 WCOL Status Flag

If the user writes the SSPBUF when a Repeated Start sequence is in progress, the WCOL flag is set and the contents of the buffer are unchanged (the write doesn't occur).

Note: Because queueing of events is not allowed, writing of the lower 5 bits of SSPCON2 is disabled until the Repeated Start condition is complete.

FIGURE 18-22: REPEATED START CONDITION WAVEFORM



18.4.10 I²C MASTER MODE TRANSMISSION

Transmission of a data byte, a 7-bit address or the other half of a 10-bit address is accomplished by simply writing a value to the SSPBUF register. This action will set the Buffer Full bit, BF, and allow the Baud Rate Generator to begin counting and start the next transmission. Each bit of address/data will be shifted out onto the SDA pin after the falling edge of SCL is asserted (see data hold time specification, parameter 106). SCL is held low for one Baud Rate Generator rollover count (TBRG). Data should be valid before SCL is released high (see data setup time specification, parameter 107). When the SCL pin is released high, it is held that way for TBRG. The data on the SDA pin must remain stable for that duration and some hold time after the next falling edge of SCL. After the eighth bit is shifted out (the falling edge of the eighth clock), the BF flag is cleared and the master releases SDA. This allows the slave device being addressed to respond with an ACK bit during the ninth bit time if an address match occurred, or if data was received properly. The status of ACK is written into the ACKDT bit on the falling edge of the ninth clock. If the master receives an Acknowledge, the Acknowledge Status bit, ACKSTAT, is cleared; if not, the bit is set. After the ninth clock, the SSPIF bit is set and the master clock (Baud Rate Generator) is suspended until the next data byte is loaded into the SSPBUF, leaving SCL low and SDA unchanged (Figure 18-23).

After the write to the SSPBUF, each bit of the address will be shifted out on the falling edge of SCL until all seven address bits and the R/W bit are completed. On the falling edge of the eighth clock, the master will deassert the SDA pin, allowing the slave to respond with an Acknowledge. On the falling edge of the ninth clock, the master will sample the SDA pin to see if the address was recognized by a slave. The status of the ACK bit is loaded into the ACKSTAT status bit (SSPCON2<6>). Following the falling edge of the ninth clock transmission of the address, the SSPIF is set, the BF flag is cleared and the Baud Rate Generator is turned off until another write to the SSPBUF takes place, holding SCL low and allowing SDA to float.

18.4.10.1 BF Status Flag

In Transmit mode, the BF bit (SSPSTAT<0>) is set when the CPU writes to SSPBUF and is cleared when all 8 bits are shifted out.

18.4.10.2 WCOL Status Flag

If the user writes to the SSPBUF when a transmit is already in progress (i.e., SSPSR is still shifting out a data byte), the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur) for 2 TcY after the SSPBUF write. If SSPBUF is rewritten within 2 TcY, the WCOL bit is set and SSPBUF is updated. This may result in a corrupted transfer. The user should verify that the WCOL is clear after each write to SSPBUF to ensure the transfer is correct. In all cases, WCOL must be cleared in software.

18.4.10.3 ACKSTAT Status Flag

In Transmit mode, the ACKSTAT bit (SSPCON2<6>) is cleared when the slave has sent an Acknowledge ($\overline{ACK} = 0$) and is set when the slave does not Acknowledge ($\overline{ACK} = 1$). A slave sends an Acknowledge when it has recognized its address (including a general call) or when the slave has properly received its data.

18.4.11 I²C MASTER MODE RECEPTION

Master mode reception is enabled by programming the Receive Enable bit, RCEN (SSPCON2<3>).

Note:	The MSSP module must be in an Idle state
	before the RCEN bit is set or the RCEN bit
	will be disregarded.

The Baud Rate Generator begins counting, and on each rollover, the state of the SCL pin changes (high-to-low/low-to-high) and data is shifted into the SSPSR. After the falling edge of the eighth clock, the receive enable flag is automatically cleared, the contents of the SSPSR are loaded into the SSPBUF, the BF flag bit is set, the SSPIF flag bit is set and the Baud Rate Generator is suspended from counting, holding SCL low. The MSSP is now in Idle state awaiting the next command. When the buffer is read by the CPU, the BF flag bit is automatically cleared. The user can then send an Acknowledge bit at the end of reception by setting the Acknowledge Sequence Enable bit, ACKEN (SSPCON2<4>).

18.4.11.1 BF Status Flag

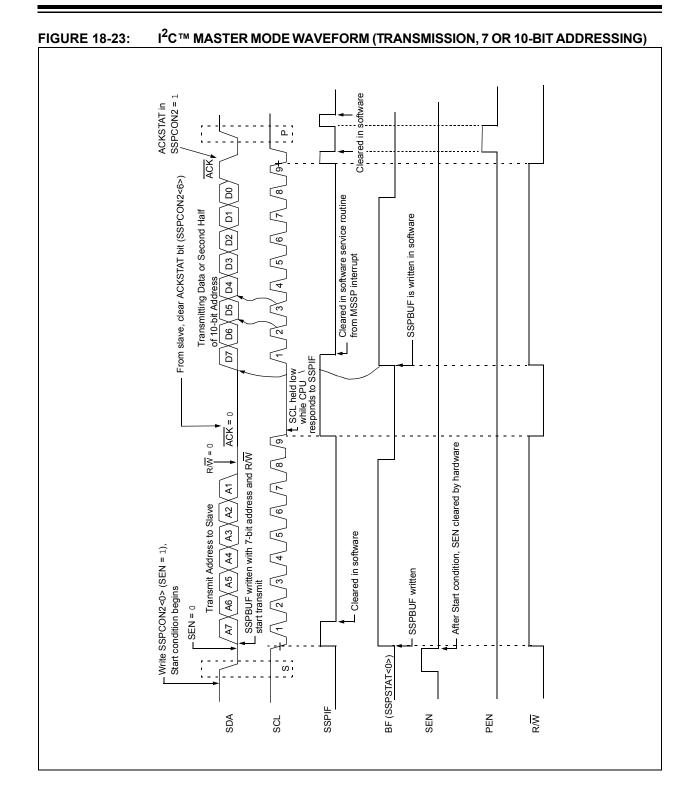
In receive operation, the BF bit is set when an address or data byte is loaded into SSPBUF from SSPSR. It is cleared when the SSPBUF register is read.

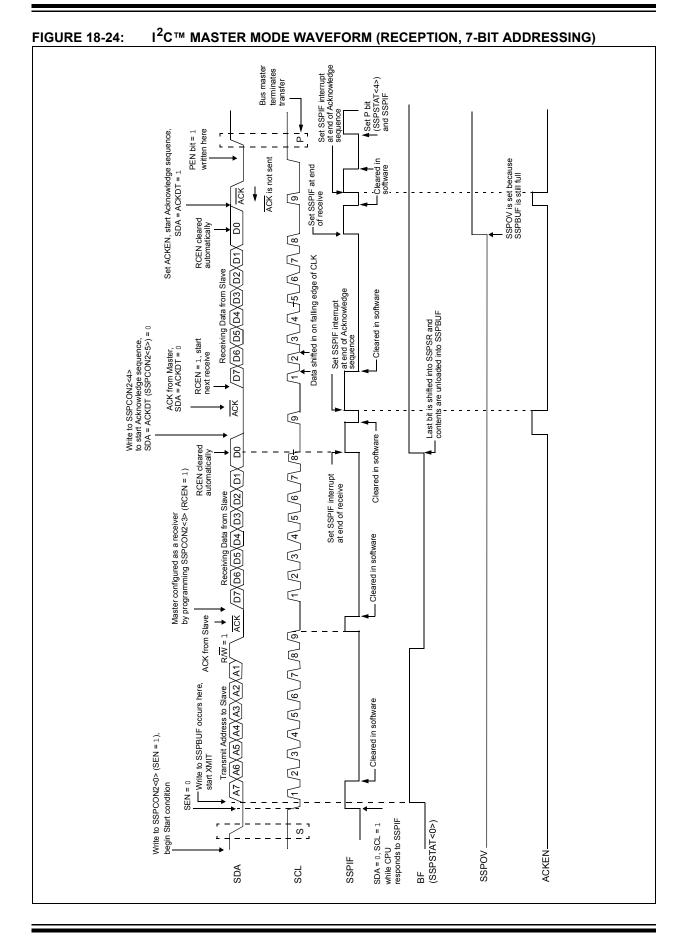
18.4.11.2 SSPOV Status Flag

In receive operation, the SSPOV bit is set when 8 bits are received into the SSPSR and the BF flag bit is already set from a previous reception.

18.4.11.3 WCOL Status Flag

If the user writes the SSPBUF when a receive is already in progress (i.e., SSPSR is still shifting in a data byte), the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur).





18.4.12 ACKNOWLEDGE SEQUENCE TIMING

An Acknowledge sequence is enabled by setting the Acknowledge Sequence Enable bit. ACKEN (SSPCON2<4>). When this bit is set, the SCL pin is pulled low and the contents of the Acknowledge data bit are presented on the SDA pin. If the user wishes to generate an Acknowledge, then the ACKDT bit should be cleared. If not, the user should set the ACKDT bit before starting an Acknowledge sequence. The Baud Rate Generator then counts for one rollover period (TBRG) and the SCL pin is deasserted (pulled high). When the SCL pin is sampled high (clock arbitration), the Baud Rate Generator counts for TBRG. The SCL pin is then pulled low. Following this, the ACKEN bit is automatically cleared, the Baud Rate Generator is turned off and the MSSP module then goes into Idle mode (Figure 18-25).

18.4.12.1 WCOL Status Flag

If the user writes the SSPBUF when an Acknowledge sequence is in progress, then WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).

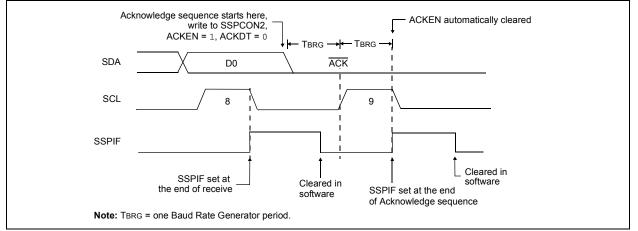
18.4.13 STOP CONDITION TIMING

A Stop bit is asserted on the SDA pin at the end of a receive/transmit by setting the Stop Sequence Enable bit, PEN (SSPCON2<2>). At the end of a receive/transmit, the SCL line is held low after the falling edge of the ninth clock. When the PEN bit is set, the master will assert the SDA line low. When the SDA line is sampled low, the Baud Rate Generator is reloaded and counts down to 0. When the Baud Rate Generator times out, the SCL pin will be brought high and one TBRG (Baud Rate Generator rollover count) later, the SDA pin will be deasserted. When the SDA pin is sampled high while SCL is high, the P bit (SSPSTAT<4>) is set. A TBRG later, the PEN bit is cleared and the SSPIF bit is set (Figure 18-26).

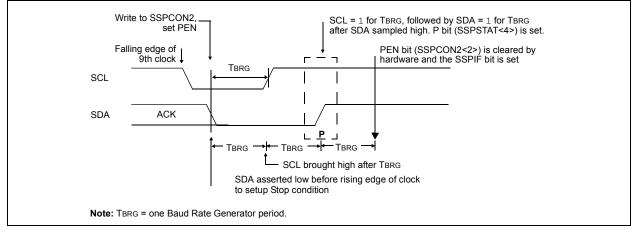
18.4.13.1 WCOL Status Flag

If the user writes the SSPBUF when a Stop sequence is in progress, then the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur).

FIGURE 18-25: ACKNOWLEDGE SEQUENCE WAVEFORM







18.4.14 SLEEP OPERATION

While in Sleep mode, the I^2C module can receive addresses or data, and when an address match or complete byte transfer occurs, wake the processor from Sleep (if the MSSP interrupt is enabled).

18.4.15 EFFECTS OF A RESET

A Reset disables the MSSP module and terminates the current transfer.

18.4.16 MULTI-MASTER MODE

In Multi-Master mode, the interrupt generation on the detection of the Start and Stop conditions allows the determination of when the bus is free. The Stop (P) and Start (S) bits are cleared from a Reset or when the MSSP module is disabled. Control of the I²C bus may be taken when the P bit (SSPSTAT<4>) is set, or the bus is Idle, with both the S and P bits clear. When the bus is busy, enabling the MSSP interrupt will generate the interrupt when the Stop condition occurs.

In multi-master operation, the SDA line must be monitored for arbitration to see if the signal level is the expected output level. This check is performed in hardware with the result placed in the BCLIF bit.

The states where arbitration can be lost are:

- · Address Transfer
- · Data Transfer
- A Start Condition
- A Repeated Start Condition
- An Acknowledge Condition

18.4.17 MULTI -MASTER COMMUNICATION, BUS COLLISION AND BUS ARBITRATION

Multi-Master mode support is achieved by bus arbitration. When the master outputs address/data bits onto the SDA pin, arbitration takes place when the master outputs a '1' on SDA by letting SDA float high, and another master asserts a '0'. When the SCL pin floats high, data should be stable. If the expected data on SDA is a '1' and the data sampled on the SDA pin = 0, then a bus collision has taken place. The master will set the Bus Collision Interrupt Flag, BCLIF, and reset the I^2C port to its Idle state (Figure 18-27).

If a transmit was in progress when the bus collision occurred, the transmission is halted, the BF flag is cleared, the SDA and SCL lines are deasserted and the SSPBUF can be written to. When the user services the bus collision Interrupt Service Routine, and if the I^2C bus is free, the user can resume communication by asserting a Start condition.

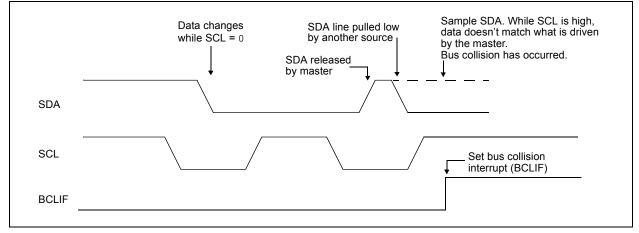
If a Start, Repeated Start, Stop or Acknowledge condition was in progress when the bus collision occurred, the condition is aborted, the SDA and SCL lines are deasserted and the respective control bits in the SSPCON2 register are cleared. When the user services the bus collision Interrupt Service Routine, and if the I²C bus is free, the user can resume communication by asserting a Start condition.

The master will continue to monitor the SDA and SCL pins. If a Stop condition occurs, the SSPIF bit will be set.

A write to the SSPBUF will start the transmission of data at the first data bit regardless of where the transmitter left off when the bus collision occurred.

In Multi-Master mode, the interrupt generation on the detection of Start and Stop conditions allows the determination of when the bus is free. Control of the I^2C bus can be taken when the P bit is set in the SSPSTAT register or the bus is Idle and the S and P bits are cleared.

FIGURE 18-27: BUS COLLISION TIMING FOR TRANSMIT AND ACKNOWLEDGE



18.4.17.1 Bus Collision During a Start Condition

During a Start condition, a bus collision occurs if:

- a) SDA or SCL are sampled low at the beginning of the Start condition (Figure 18-28).
- b) SCL is sampled low before SDA is asserted low (Figure 18-29).

During a Start condition, both the SDA and the SCL pins are monitored.

If the SDA pin is already low, or the SCL pin is already low, then all of the following occur:

- the Start condition is aborted;
- the BCLIF flag is set; and
- the MSSP module is reset to its Idle state (Figure 18-28).

The Start condition begins with the SDA and SCL pins deasserted. When the SDA pin is sampled high, the Baud Rate Generator is loaded from SSPADD<6:0> and counts down to 0. If the SCL pin is sampled low while SDA is high, a bus collision occurs, because it is assumed that another master is attempting to drive a data '1' during the Start condition.

If the SDA pin is sampled low during this count, the BRG is reset and the SDA line is asserted early (Figure 18-30). If, however, a '1' is sampled on the SDA pin, the SDA pin is asserted low at the end of the BRG count. The Baud Rate Generator is then reloaded and counts down to 0. If the SCL pin is sampled as '0' during this time, a bus collision does not occur. At the end of the BRG count, the SCL pin is asserted low.

Note: The reason that bus collision is not a factor during a Start condition is that no two bus masters can assert a Start condition at the exact same time. Therefore, one master will always assert SDA before the other. This condition does not cause a bus collision because the two masters must be allowed to arbitrate the first address following the Start condition. If the address is the same, arbitration must be allowed to continue into the data portion, Repeated Start or Stop conditions.

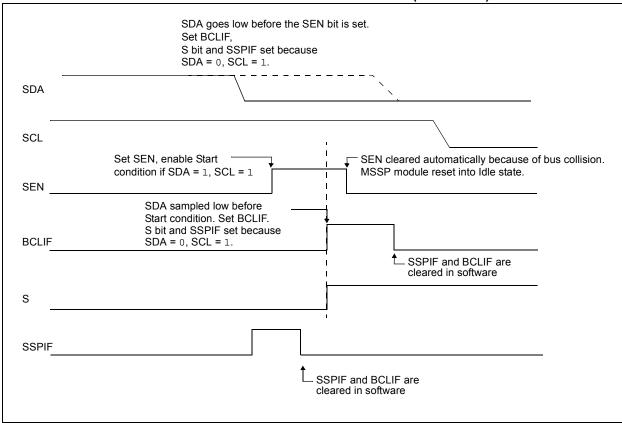
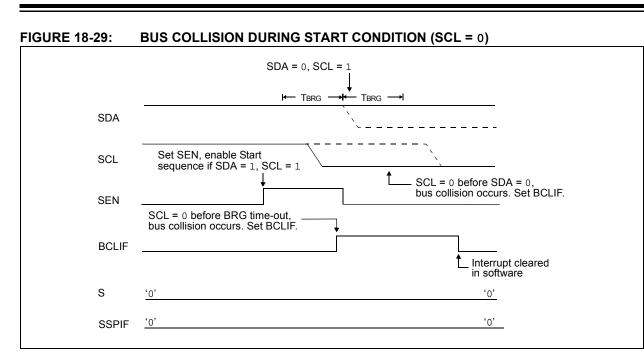
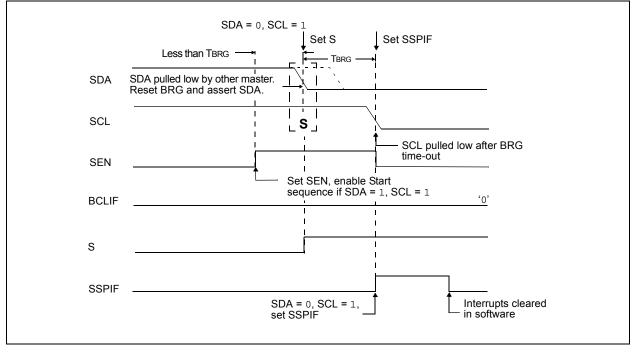


FIGURE 18-28: BUS COLLISION DURING START CONDITION (SDA ONLY)







18.4.17.2 Bus Collision During a Repeated Start Condition

During a Repeated Start condition, a bus collision occurs if:

- a) A low level is sampled on SDA when SCL goes from a low level to a high level.
- SCL goes low before SDA is asserted low, indicating that another master is attempting to transmit a data '1'.

When the user deasserts SDA, and the pin is allowed to float high, the BRG is loaded with SSPADD<6:0> and counts down to 0. The SCL pin is then deasserted, and when sampled high, the SDA pin is sampled.

If SDA is low, a bus collision has occurred (i.e., another master is attempting to transmit a data '0'; see Figure 18-31). If SDA is sampled high, the BRG is reloaded and begins counting. If SDA goes from high-to-low before the BRG times out, no bus collision occurs because no two masters can assert SDA at exactly the same time.

If SCL goes from high-to-low before the BRG times out, and SDA has not already been asserted, a bus collision occurs. In this case, another master is attempting to transmit a data '1' during the Repeated Start condition (see Figure 18-32).

If, at the end of the BRG time-out, both SCL and SDA are still high, the SDA pin is driven low and the BRG is reloaded and begins counting. At the end of the count, regardless of the status of the SCL pin, the SCL pin is driven low and the Repeated Start condition is complete.

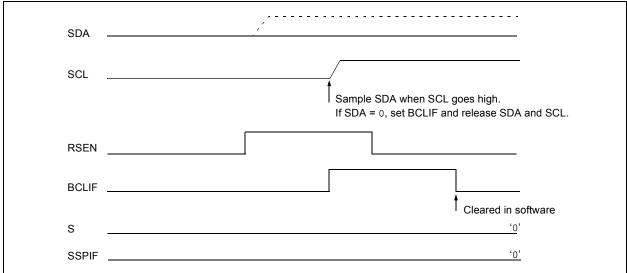
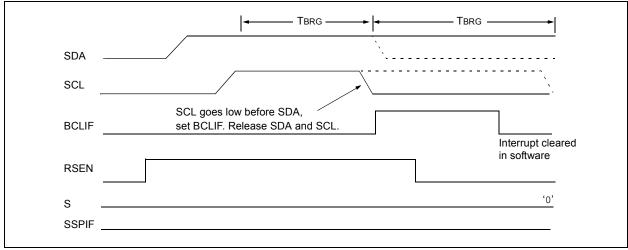


FIGURE 18-31: BUS COLLISION DURING A REPEATED START CONDITION (CASE 1)

FIGURE 18-32: BUS COLLISION DURING REPEATED START CONDITION (CASE 2)



18.4.17.3 Bus Collision During a Stop Condition

Bus collision occurs during a Stop condition if:

- a) After the SDA pin has been deasserted and allowed to float high, SDA is sampled low after the BRG has timed out.
- b) After the SCL pin is deasserted, SCL is sampled low before SDA goes high.

The Stop condition begins with SDA asserted low. When SDA is sampled low, the SCL pin is allowed to float. When the pin is sampled high (clock arbitration), the Baud Rate Generator is loaded with SSPADD<6:0> and counts down to 0. After the BRG times out, SDA is sampled. If SDA is sampled low, a bus collision has occurred. This is due to another master attempting to drive a data '0' (Figure 18-33). If the SCL pin is sampled low before SDA is allowed to float high, a bus collision occurs. This is another case of another master attempting to drive a data '0' (Figure 18-34).

FIGURE 18-33: BUS COLLISION DURING A STOP CONDITION (CASE 1)

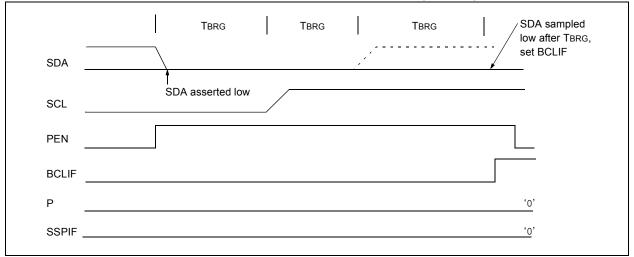
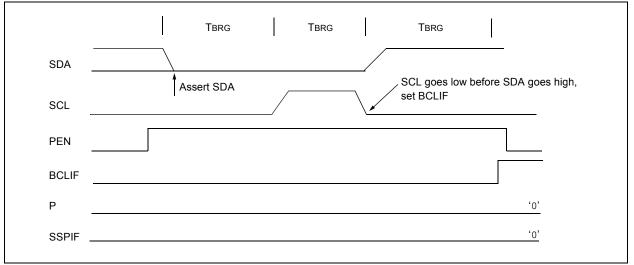


FIGURE 18-34: BUS COLLISION DURING A STOP CONDITION (CASE 2)



PIC18F87J90 FAMILY

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59
PIR1	—	ADIF	RC1IF	TX1IF	SSPIF	_	TMR2IF	TMR1IF	62
PIE1	—	ADIE	RC1IE	TX1IE	SSPIE	_	TMR2IE	TMR1IE	62
IPR1	_	ADIP	RC1IP	TX1IP	SSPIP		TMR2IP	TMR1IP	62
PIR2	OSCFIF	CMIF	_	_	BCLIF	LVDIF	TMR3IF	_	62
PIE2	OSCFIE	CMIE	—	_	BCLIE	LVDIE	TMR3IE	—	62
IPR2	OSCFIP	CMIP	_	_	BCLIP	LVDIP	TMR3IP	_	62
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	62
SSPBUF	MSSP Red	eive Buffer/	Transmit Re	gister					60
SSPADD	MSSP Add MSSP Bau	lress Regist Id Rate Relo	er (I ² C™ Sla bad Register	ave mode), (I ² C Master	mode)				60
SSPCON1	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	60
SSPCON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	60
	GCEN	ACKSTAT	ADMSK5 ⁽¹⁾	ADMSK4 ⁽¹⁾	ADMSK3 ⁽¹⁾	ADMSK2 ⁽¹⁾	ADMSK1 ⁽¹⁾	SEN	00
SSPSTAT	SMP	CKE	D/Ā	Р	S	R/W	UA	BF	60

TABLE 18-4: REGISTERS ASSOCIATED WITH I²C[™] OPERATION

Legend: – = unimplemented, read as '0'. Shaded cells are not used by the MSSP module in I^2C^{TM} mode.

Note 1: Alternate bit definitions for use in I²C Slave mode operations only.

19.0 ENHANCED UNIVERSAL SYNCHRONOUS ASYNCHRONOUS RECEIVER TRANSMITTER (EUSART)

PIC18F87J90 family devices have three serial I/O modules: the MSSP module, discussed in the previous chapter and two Universal Synchronous Asynchronous Receiver Transmitter (USART) modules. (Generically, the USART is also known as a Serial Communications Interface or SCI.) The USART can be configured as a full-duplex, asynchronous system that can communicate with peripheral devices, such as CRT terminals and personal computers. It can also be configured as a half-duplex synchronous system that can communicate with peripheral devices, such as A/D or D/A integrated circuits, serial EEPROMs, etc.

There are two distinct implementations of the USART module in these devices: the Enhanced USART (EUSART) discussed here and the Addressable USART discussed in the next chapter. For this device family, USART1 always refers to the EUSART, while USART2 is always the AUSART.

The EUSART and AUSART modules implement the same core features for serial communications; their basic operation is essentially the same. The EUSART module provides additional features, including Automatic Baud Rate Detection and calibration, automatic wake-up on Sync Break reception, and 12-bit Break character transmit. These features make it ideally suited for use in Local Interconnect Network bus (LIN/J2602 bus) systems.

The EUSART can be configured in the following modes:

- Asynchronous (full-duplex) with:
 - Auto-wake-up on character reception
 - Auto-baud calibration
 - 12-bit Break character transmission
- Synchronous Master (half-duplex) with selectable clock polarity
- Synchronous Slave (half-duplex) with selectable clock polarity

The pins of the EUSART are multiplexed with the functions of PORTC (RC6/TX1/CK1/SEG27 and RC7/RX1/DT1/SEG28). In order to configure these pins as an EUSART:

- bit, SPEN (RCSTA1<7>), must be set (= 1)
- bit, TRISC<7>, must be set (= 1)
- bit, TRISC<6>, must be set (= 1)
- Note: The EUSART control will automatically reconfigure the pin from input to output as needed.

The driver for the TX1 output pin can also be optionally configured as an open-drain output. This feature allows the voltage level on the pin to be pulled to a higher level through an external pull-up resistor, and allows the output to communicate with external circuits without the need for additional level shifters.

The open-drain output option is controlled by the U1OD bit (LATG<6>). Setting the bit configures the pin for open-drain operation.

19.1 Control Registers

The operation of the Enhanced USART module is controlled through three registers:

- Transmit Status and Control Register 1 (TXSTA1)
- Receive Status and Control Register 1 (RCSTA1)
- Baud Rate Control Register 1 (BAUDCON1)

The registers are described in Register 19-1, Register 19-2 and Register 19-3.

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R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-1	R/W-0
CSRC	TX9	TXEN ⁽¹⁾	SYNC	SENDB	BRGH	TRMT	TX9D
bit 7	I						bit
Legend:							
R = Reada		W = Writable		-	nented bit, read		
-n = Value	at POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown
bit 7	CSRC: Cloc	k Source Select	bit				
	<u>Asynchrono</u> Don't care.	<u>us mode:</u>					
		<u>is mode:</u> mode (clock gen iode (clock from					
bit 6	TX9: 9-Bit T	ransmit Enable	bit				
		9-bit transmissi 8-bit transmissi					
bit 5	TXEN: Tran 1 = Transm 0 = Transm)				
bit 4		SART Mode Sele	ct bit				
	1 = Synchro 0 = Asynch	onous mode ronous mode					
bit 3	SENDB: Se	nd Break Chara	cter bit				
	<u>Asynchrono</u>	<u>us mode:</u>					
		ync Break on ne reak transmissio		n (cleared by ha	ardware upon o	completion)	
	<u>Synchronou</u> Don't care.	<u>is mode:</u>					
bit 2	BRGH: High	n Baud Rate Sel	ect bit				
	Asynchrono 1 = High sp 0 = Low sp	eed					
	Synchronou Unused in th	s mode:					
bit 1	TRMT: Tran	smit Shift Regist	er Status bit				
	1 = TSR en 0 = TSR ful						
bit 0	TX9D: 9th b						
		oit of Transmit Da	ata				

REGISTER 19-1: TXSTA1: EUSART TRANSMIT STATUS AND CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-0	R-0	R-x
SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D
oit 7							bit (
_egend:							
R = Readab		W = Writable b	it	U = Unimplem		ıd as '0'	
-n = Value a	t POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	iown
bit 7	SDEN: Soria	al Port Enable bit					
		ort enabled (conf	iaures RX1/C	T1 and TX1/CK	1 pins as seri	al port pins)	
		ort disabled (held			. p	()	
bit 6	RX9: 9-Bit R	eceive Enable bi	t				
		9-bit reception 8-bit reception					
bit 5	SREN: Sing	le Receive Enabl	e bit				
	<u>Asynchronou</u> Don't care.	<u>us mode</u> :					
		s mode – Master: s single receive					
		s single receive eared after recept	tion is comple	ete.			
	Synchronous Don't care.	<u>s mode – Slave:</u>					
bit 4	CREN: Cont	inuous Receive I	Enable bit				
	Asynchronou 1 = Enables	receiver					
	0 = Disables						
		s continuous rece s continuous rece		le bit, CREN, is	cleared (CRE	EN overrides SR	EN)
bit 3		dress Detect Ena	-				
	Asynchronou	us mode 9-bit (R)	(9 = <u>1)</u> :				
		address detection					
		s address detecti us mode 9-bit (R)		are received and	the ninth bit	can be used as	a parity bit
	Don't care.		<u>(9 – 0)</u> .				
bit 2	FERR: Fram	ning Error bit					
		error (can be up	dated by rea	ding RCREG1 re	egister and re	ceiving next val	d byte)
bit 1	OERR: Over	•					
	1 = Overrun 0 = No over	i error (can be cle run error	ared by clea	ring bit, CREN)			
oit 0		it of Received Da	ita				
	This can be	address/data bit	or a parity hit	and must be ca	loulated by us	or firmwara	

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R/W-0	R-1	R/W - 0	R/W-0	R/W-0	U-0	R/W-0	R/W-0
ABDOVF	RCIDL	RXDTP	TXCKP	BRG16		WUE	ABDEN
bit 7							bit (
Legend:							
R = Readable		W = Writable		-	mented bit, rea		
-n = Value at	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unki	nown
bit 7	ABDOVF: Au	to-Baud Acqui	sition Rollover	Status bit			
		ollover has occ rollover has oc		uto-Baud Rate	Detect mode (must be cleare	d in software)
bit 6	RCIDL: Rece	ive Operation I	dle Status bit				
		operation is Idl operation is ac					
bit 5	RXDTP: Rece	eived Data Pola	arity Select bit				
	Asynchronous 1 = RXx data 0 = RXx data		I				
	Synchronous 1 = CKx cloc						
bit 4	TXCKP: Cloc	k and Data Pol	arity Select bit				
	Asynchronous 1 = TXx data 0 = TXx data						
		<u>mode:</u> ks are inverted ks are not inve					
bit 3	BRG16: 16-B	it Baud Rate R	egister Enable	e bit			
				H1 and SPBR		BRGH1 value i	gnored
bit 2	Unimplemen	ted: Read as '	0'				
bit 1	WUE: Wake-u	up Enable bit					
	hardware		ising edge		rupt generated	on falling edge	e; bit cleared i
	Synchronous Unused in this						
bit 0	ABDEN: Auto	-Baud Detect	Enable bit				
	cleared ir 0 = Baud rate	aud rate meas n hardware upo e measuremen	on completion.		er. Requires re	eception of a Sy	ync field (55h
	Synchronous Unused in this						

19.2 EUSART Baud Rate Generator (BRG)

The BRG is a dedicated, 8-bit or 16-bit generator that supports both the Asynchronous and Synchronous modes of the EUSART. By default, the BRG operates in 8-bit mode; setting the BRG16 bit (BAUDCON1<3>) selects 16-bit mode.

The SPBRGH1:SPBRG1 register pair controls the period of a free-running timer. In Asynchronous mode, the BRGH (TXSTA1<2>) and BRG16 (BAUDCON1<3>) bits also control the baud rate. In Synchronous mode, BRGH is ignored. Table 19-1 shows the formula for computation of the baud rate for different EUSART modes that only apply in Master mode (internally generated clock).

Given the desired baud rate and Fosc, the nearest integer value for the SPBRGH1:SPBRG1 registers can be calculated using the formulas in Table 19-1. From this, the error in baud rate can be determined. An example calculation is shown in Example 19-1. Typical baud rates and error values for the various Asynchronous modes are shown in Table 19-2. It may be advantageous to use the high baud rate (BRGH = 1) or the 16-bit BRG to reduce the baud rate error, or achieve a slow baud rate for a fast oscillator frequency.

Writing a new value to the SPBRGH1:SPBRG1 registers causes the BRG timer to be reset (or cleared). This ensures the BRG does not wait for a timer overflow before outputting the new baud rate.

19.2.1 OPERATION IN POWER-MANAGED MODES

The device clock is used to generate the desired baud rate. When one of the power-managed modes is entered, the new clock source may be operating at a different frequency. This may require an adjustment to the value in the SPBRG1 register pair.

19.2.2 SAMPLING

The data on the RX1 pin is sampled three times by a majority detect circuit to determine if a high or a low level is present at the RX1 pin.

C	onfiguration B	its		David Data Farmada		
SYNC	NC BRG16 BRGH		BRG/EUSART Mode	Baud Rate Formula		
0	0	0	8-bit/Asynchronous	Fosc/[64 (n + 1)]		
0	0	1	8-bit/Asynchronous			
0	1	0	16-bit/Asynchronous	Fosc/[16 (n + 1)]		
0	1	1	16-bit/Asynchronous			
1	0	x	8-bit/Synchronous	Fosc/[4 (n + 1)]		
1	1	x	16-bit/Synchronous			

TABLE 19-1: BAUD RATE FORMULAS

Legend: x = Don't care, n = Value of SPBRGH1:SPBRG1 register pair

EXAMPLE 19-1: CALCULATING BAUD RATE ERROR

For a device with FOSC of 16 MHz, desired baud rate of 9600, Asynchronous mode, 8-bit BRG:
Desired Baud Rate = $FOSC/(64 ([SPBRGH1:SPBRG1] + 1))$
Solving for SPBRGH1:SPBRG1:
X = ((Fosc/Desired Baud Rate)/64) - 1
= ((1600000/9600)/64) - 1
= [25.042] = 25
Calculated Baud Rate = $16000000/(64 (25 + 1))$
= 9615
Error = (Calculated Baud Rate – Desired Baud Rate)/Desired Baud Rate
= (9615 - 9600)/9600 = 0.16%

TABLE 19-2: REGISTERS ASSOCIATED WITH THE BAUD RATE GENERATOR

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	61
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	61
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN	63
SPBRGH1	EUSART E	Baud Rate	Generato	Register	High Byte				63
SPBRG1		61							
Legend: — = unimplemented, read as '0'. Shaded cells are not used by the BRG.									

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					SYNC	= 0, BRGH	I = 0, BRG	616 = 0				
BAUD RATE	Fosc	= 40.000) MHz	Fosc = 20.000 MHz			Foso	: = 10.000) MHz	Fosc = 8.000 MHz		
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	—	_	_	_	_	_	_	_	_	_	_	_
1.2	—	—	_	1.221	1.73	255	1.202	0.16	129	1.201	-0.16	103
2.4	2.441	1.73	255	2.404	0.16	129	2.404	0.16	64	2.403	-0.16	51
9.6	9.615	0.16	64	9.766	1.73	31	9.766	1.73	15	9.615	-0.16	12
19.2	19.531	1.73	31	19.531	1.73	15	19.531	1.73	7	_	_	_
57.6	56.818	-1.36	10	62.500	8.51	4	52.083	-9.58	2	—	_	_
115.2	125.000	8.51	4	104.167	-9.58	2	78.125	-32.18	1	—	_	_

TABLE 19-3: BAUD RATES FOR ASYNCHRONOUS MODES

		SYNC = 0, BRGH = 0, BRG16 = 0											
BAUD	Fos	c = 4.000	MHz	Fos	c = 2.000	MHz	Fosc = 1.000 MHz						
RATE (K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)				
0.3	0.300	0.16	207	0.300	-0.16	103	0.300	-0.16	51				
1.2	1.202	0.16	51	1.201	-0.16	25	1.201	-0.16	12				
2.4	2.404	0.16	25	2.403	-0.16	12	—	_	_				
9.6	8.929	-6.99	6	—	_	_	—	_	_				
19.2	20.833	8.51	2	—	_	_	—	_	_				
57.6	62.500	8.51	0	—	_	_	—	_	_				
115.2	62.500	-45.75	0	_	—	_	_		_				

					SYNC	= 0, BRGH	i = 1, BRG	i 16 = 0					
BAUD RATE	Fosc	Fosc = 40.000 MHz			Fosc = 20.000 MHz			Fosc = 10.000 MHz			Fosc = 8.000 MHz		
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	
0.3	—	_	_	—	_	_	_	_	_	_	_	_	
1.2	—	—	—	—	—	—	—	—	—	—	—	—	
2.4	—	_	_	—	_	_	2.441	1.73	255	2.403	-0.16	207	
9.6	9.766	1.73	255	9.615	0.16	129	9.615	0.16	64	9.615	-0.16	51	
19.2	19.231	0.16	129	19.231	0.16	64	19.531	1.73	31	19.230	-0.16	25	
57.6	58.140	0.94	42	56.818	-1.36	21	56.818	-1.36	10	55.555	3.55	8	
115.2	113.636	-1.36	21	113.636	-1.36	10	125.000	8.51	4	_	—	—	

			S	YNC = 0, E	BRGH = 1	, BRG16 =	0			
BAUD	Foso	= 4.000	MHz	Fos	c = 2.000	MHz	Fos	Fosc = 1.000 MHz		
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	
0.3	_		_	_	_	_	0.300	-0.16	207	
1.2	1.202	0.16	207	1.201	-0.16	103	1.201	-0.16	51	
2.4	2.404	0.16	103	2.403	-0.16	51	2.403	-0.16	25	
9.6	9.615	0.16	25	9.615	-0.16	12	_	_	_	
19.2	19.231	0.16	12	—	—	—	—	—	—	
57.6	62.500	8.51	3	—	_	—	—	_	_	
115.2	125.000	8.51	1	_	_	—	_	_	—	

				SYNC = 0, BRGH = 0, BRG16 = 1											
BAUD	Fosc	Fosc = 40.000 MHz			Fosc = 20.000 MHz			Fosc = 10.000 MHz			Fosc = 8.000 MHz				
RATE (K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)			
0.3	0.300	0.00	8332	0.300	0.02	4165	0.300	0.02	2082	0.300	-0.04	1665			
1.2	1.200	0.02	2082	1.200	-0.03	1041	1.200	-0.03	520	1.201	-0.16	415			
2.4	2.402	0.06	1040	2.399	-0.03	520	2.404	0.16	259	2.403	-0.16	207			
9.6	9.615	0.16	259	9.615	0.16	129	9.615	0.16	64	9.615	-0.16	51			
19.2	19.231	0.16	129	19.231	0.16	64	19.531	1.73	31	19.230	-0.16	25			
57.6	58.140	0.94	42	56.818	-1.36	21	56.818	-1.36	10	55.555	3.55	8			
115.2	113.636	-1.36	21	113.636	-1.36	10	125.000	8.51	4	—	_	_			

TABLE 19-3:	BAUD RATES FOR	ASYNCHRONOUS M	ODES (CONTINUED)

		SYNC = 0, BRGH = 0, BRG16 = 1											
BAUD	Fosc = 4.000 MHz			Fos	c = 2.000	MHz	Fos	c = 1.000	MHz				
RATE (K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)				
0.3	0.300	0.04	832	0.300	-0.16	415	0.300	-0.16	207				
1.2	1.202	0.16	207	1.201	-0.16	103	1.201	-0.16	51				
2.4	2.404	0.16	103	2.403	-0.16	51	2.403	-0.16	25				
9.6	9.615	0.16	25	9.615	-0.16	12	_	_	_				
19.2	19.231	0.16	12	_	_	_	_	_	_				
57.6	62.500	8.51	3	—	_	_	—	_	_				
115.2	125.000	8.51	1	_	—	—	_	—					

		SYNC = 0, BRGH = 1, BRG16 = 1 or SYNC = 1, BRG16 = 1														
BAUD RATE	Fosc	= 40.000) MHz	Fosc = 20.000 MHz			Fosc	= 10.000) MHz	Fos	c = 8.000	MHz				
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)				
0.3	0.300	0.00	33332	0.300	0.00	16665	0.300	0.00	8332	0.300	-0.01	6665				
1.2	1.200	0.00	8332	1.200	0.02	4165	1.200	0.02	2082	1.200	-0.04	1665				
2.4	2.400	0.02	4165	2.400	0.02	2082	2.402	0.06	1040	2.400	-0.04	832				
9.6	9.606	0.06	1040	9.596	-0.03	520	9.615	0.16	259	9.615	-0.16	207				
19.2	19.193	-0.03	520	19.231	0.16	259	19.231	0.16	129	19.230	-0.16	103				
57.6	57.803	0.35	172	57.471	-0.22	86	58.140	0.94	42	57.142	0.79	34				
115.2	114.943	-0.22	86	116.279	0.94	42	113.636	-1.36	21	117.647	-2.12	16				

		SYN	IC = 0, BRO	GH = 1, BF	RG16 = 1	or SYNC =	: 1, BRG1	6=1	
BAUD RATE	Foso	c = 4.000	MHz	Fos	c = 2.000	MHz	Fos	c = 1.000	MHz
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	0.300	0.01	3332	0.300	-0.04	1665	0.300	-0.04	832
1.2	1.200	0.04	832	1.201	-0.16	415	1.201	-0.16	207
2.4	2.404	0.16	415	2.403	-0.16	207	2.403	-0.16	103
9.6	9.615	0.16	103	9.615	-0.16	51	9.615	-0.16	25
19.2	19.231	0.16	51	19.230	-0.16	25	19.230	-0.16	12
57.6	58.824	2.12	16	55.555	3.55	8	—	_	—
115.2	111.111	-3.55	8	—	_	_	—	_	—

19.2.3 AUTO-BAUD RATE DETECT

The Enhanced USART module supports the automatic detection and calibration of baud rate. This feature is active only in Asynchronous mode and while the WUE bit is clear.

The automatic baud rate measurement sequence (Figure 19-1) begins whenever a Start bit is received and the ABDEN bit is set. The calculation is self-averaging.

In the Auto-Baud Rate Detect (ABD) mode, the clock to the BRG is reversed. Rather than the BRG clocking the incoming RX1 signal, the RX1 signal is timing the BRG. In ABD mode, the internal Baud Rate Generator is used as a counter to time the bit period of the incoming serial byte stream.

Once the ABDEN bit is set, the state machine will clear the BRG and look for a Start bit. The Auto-Baud Rate Detect must receive a byte with the value, 55h (ASCII "U", which is also the LIN/J2602 bus Sync character), in order to calculate the proper bit rate. The measurement is taken over both a low and a high bit time in order to minimize any effects caused by asymmetry of the incoming signal. After a Start bit, the SPBRG1 begins counting up, using the preselected clock source on the first rising edge of RX1. After eight bits on the RX1 pin, or the fifth rising edge, an accumulated value totalling the proper BRG period is left in the SPBRGH1:SPBRG1 register pair. Once the 5th edge is seen (this should correspond to the Stop bit), the ABDEN bit is automatically cleared.

If a rollover of the BRG occurs (an overflow from FFFFh to 0000h), the event is trapped by the ABDOVF status bit (BAUDCON1<7>). It is set in hardware by BRG rollovers and can be set or cleared by the user in software. ABD mode remains active after rollover events and the ABDEN bit remains set (Figure 19-2).

While calibrating the baud rate period, the BRG registers are clocked at 1/8th the preconfigured clock rate. Note that the BRG clock can be configured by the BRG16 and BRGH bits. The BRG16 bit must be set to use both SPBRG1 and SPBRGH1 as a 16-bit counter. This allows the user to verify that no carry occurred for 8-bit modes by checking for 00h in the SPBRGH1 register. Refer to Table 19-4 for counter clock rates to the BRG.

While the ABD sequence takes place, the EUSART state machine is held in Idle. The RC1IF interrupt is set once the fifth rising edge on RX1 is detected. The value in the RCREG1 needs to be read to clear the RC1IF interrupt. The contents of RCREG1 should be discarded.

- Note 1: If the WUE bit is set with the ABDEN bit, Auto-Baud Rate Detection will occur on the byte *following* the Break character.
 - 2: It is up to the user to determine that the incoming character baud rate is within the range of the selected BRG clock source. Some combinations of oscillator frequency and EUSART baud rates are not possible due to bit error rates. Overall system timing and communication baud rates must be taken into consideration when using the Auto-Baud Rate Detection feature.

TABLE 19-4: BRG COUNTER CLOCK RATES

BRG16	BRGH	BRG Counter Clock
0	0	Fosc/512
0	1	Fosc/128
1	0	Fosc/128
1	1	Fosc/32

Note: During the ABD sequence, SPBRG1 and SPBRGH1 are both used as a 16-bit counter, independent of the BRG16 setting.

19.2.3.1 ABD and EUSART Transmission

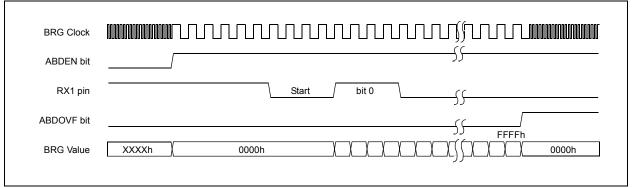
Since the BRG clock is reversed during ABD acquisition, the EUSART transmitter cannot be used during ABD. This means that whenever the ABDEN bit is set, TXREG1 cannot be written to. Users should also ensure that ABDEN does not become set during a transmit sequence. Failing to do this may result in unpredictable EUSART operation.

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GURE 19	-1. AUIC	MIATIC BAUD	RATE CALCULATION	
BRG Value	XXXXh X	0000h		001Ch
RX1 pin		Start	Edge #1	Edge #5
BRG Clock	הטונטטטטטטטעטטעטטעטטטטטטטין דר	www.w		
ABDEN bit	Set by User			Auto-Cleared
RC1IF bit (Interrupt)				
Read RCREG1				
SPBRG1			XXXXh	χ 1Ch
SPBRGH1			XXXXh) 00h

FIGURE 19-1: AUTOMATIC BAUD RATE CALCULATION

FIGURE 19-2: BRG OVERFLOW SEQUENCE



19.3 EUSART Asynchronous Mode

The Asynchronous mode of operation is selected by clearing the SYNC bit (TXSTA1<4>). In this mode, the EUSART uses standard Non-Return-to-Zero (NRZ) format (one Start bit, eight or nine data bits and one Stop bit). The most common data format is 8 bits. An on-chip dedicated 8-bit/16-bit Baud Rate Generator can be used to derive standard baud rate frequencies from the oscillator.

The EUSART transmits and receives the LSb first. The EUSART's transmitter and receiver are functionally independent, but use the same data format and baud rate. The Baud Rate Generator produces a clock, either x16 or x64 of the bit shift rate, depending on the BRGH and BRG16 bits (TXSTA1<2> and BAUDCON1<3>). Parity is not supported by the hardware but can be implemented in software and stored as the 9th data bit.

When operating in Asynchronous mode, the EUSART module consists of the following important elements:

- Baud Rate Generator
- · Sampling Circuit
- Asynchronous Transmitter
- · Asynchronous Receiver
- Auto-Wake-up on Sync Break Character
- 12-Bit Break Character Transmit
- Auto-Baud Rate Detection

19.3.1 EUSART ASYNCHRONOUS TRANSMITTER

The EUSART transmitter block diagram is shown in Figure 19-3. The heart of the transmitter is the Transmit (Serial) Shift Register (TSR). The Shift register obtains its data from the Read/Write Transmit Buffer Register, TXREG1. The TXREG1 register is loaded with data in software. The TSR register is not loaded until the Stop bit has been transmitted from the previous load. As soon as the Stop bit is transmitted, the TSR is loaded with new data from the TXREG1 register (if available).

Once the TXREG1 register transfers the data to the TSR register (occurs in one TCY), the TXREG1 register is empty and the TX1IF flag bit (PIR1<4>) is set. This interrupt can be enabled or disabled by setting or clearing the interrupt enable bit, TX1IE (PIE1<4>). TX1IF will be set regardless of the state of TX1IE; it cannot be cleared in software. TX1IF is also not cleared immediately upon loading TXREG1, but becomes valid in the second instruction cycle following the load instruction. Polling TX1IF immediately following a load of TXREG1 will return invalid results.

While TX1IF indicates the status of the TXREG1 register, another bit, TRMT (TXSTA1<1>), shows the status of the TSR register. TRMT is a read-only bit which is set when the TSR register is empty. No interrupt logic is tied to this bit, so the user has to poll this bit in order to determine if the TSR register is empty.

Note 1:	The TSR register is not mapped in data memory, so it is not available to the user.
2:	Flag bit, TX1IF, is set when enable bit, TXEN, is set.

To set up an Asynchronous Transmission:

- 1. Initialize the SPBRGH1:SPBRG1 registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
- 2. Enable the asynchronous serial port by clearing bit, SYNC, and setting bit, SPEN.
- 3. If interrupts are desired, set enable bit, TX1IE.
- 4. If 9-bit transmission is desired, set transmit bit, TX9; can be used as address/data bit.
- 5. Enable the transmission by setting bit, TXEN, which will also set bit, TX1IF.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit, TX9D.
- 7. Load data to the TXREG1 register (starts transmission).
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

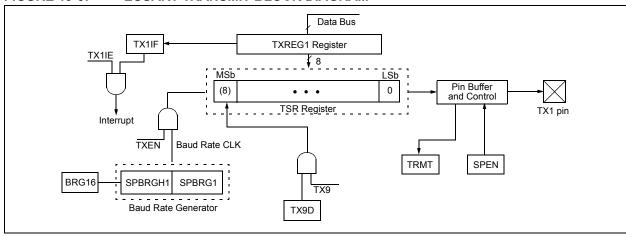


FIGURE 19-3: EUSART TRANSMIT BLOCK DIAGRAM

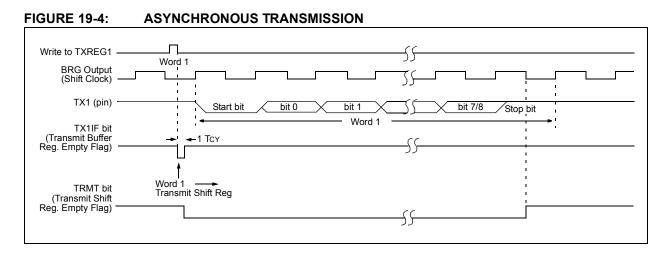


FIGURE 19-5: ASYNCHRONOUS TRANSMISSION (BACK TO BACK)

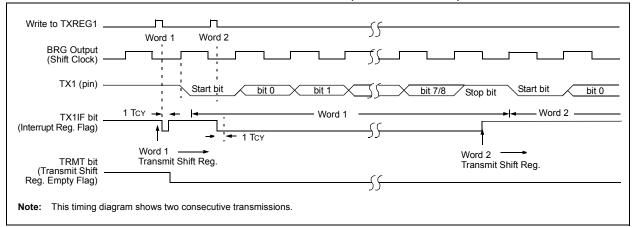


TABLE 19-5: REGISTERS ASSOCIATED WITH ASYNCHRONOUS TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page	
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59	
PIR1	—	ADIF	RC1IF	TX1IF	SSPIF	_	TMR2IF	TMR1IF	62	
PIE1	_	ADIE	RC1IE	TX1IE	SSPIE	_	TMR2IE	TMR1IE	62	
IPR1	—	ADIP	RC1IP	TX1IP	SSPIP	_	TMR2IP	TMR1IP	62	
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	61	
TXREG1	EUSART T	ransmit Reg	ister						61	
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	61	
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	—	WUE	ABDEN	63	
SPBRGH1	EUSART B	aud Rate G	enerator Re	gister High	Byte				63	
SPBRG1	EUSART B	USART Baud Rate Generator Register Low Byte								
LATG	U2OD	U10D	_	LATG4	LATG3	LATG2	LATG1	LATG0	62	

Legend: — = unimplemented locations read as '0'. Shaded cells are not used for asynchronous transmission.

19.3.2 EUSART ASYNCHRONOUS RECEIVER

The receiver block diagram is shown in Figure 19-6. The data is received on the RX1 pin and drives the data recovery block. The data recovery block is actually a high-speed shifter, operating at x16 times the baud rate, whereas the main receive serial shifter operates at the bit rate or at Fosc. This mode would typically be used in RS-232 systems.

To set up an Asynchronous Reception:

- 1. Initialize the SPBRGH1:SPBRG1 registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
- 2. Enable the asynchronous serial port by clearing bit, SYNC, and setting bit, SPEN.
- 3. If interrupts are desired, set enable bit, RC1IE.
- 4. If 9-bit reception is desired, set bit, RX9.
- 5. Enable the reception by setting bit, CREN.
- Flag bit, RC1IF, will be set when reception is complete and an interrupt will be generated if enable bit, RC1IE, was set.
- 7. Read the RCSTA1 register to get the 9th bit (if enabled) and determine if any error occurred during reception.
- 8. Read the 8-bit received data by reading the RCREG1 register.
- 9. If any error occurred, clear the error by clearing enable bit, CREN.
- 10. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

19.3.3 SETTING UP 9-BIT MODE WITH ADDRESS DETECT

This mode would typically be used in RS-485 systems. To set up an Asynchronous Reception with Address Detect Enable:

- 1. Initialize the SPBRGH1:SPBRG1 registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
- 2. Enable the asynchronous serial port by clearing the SYNC bit and setting the SPEN bit.
- If interrupts are required, set the RCEN bit and select the desired priority level with the RC1IP bit.
- 4. Set the RX9 bit to enable 9-bit reception.
- 5. Set the ADDEN bit to enable address detect.
- 6. Enable reception by setting the CREN bit.
- The RC1IF bit will be set when reception is complete. The interrupt will be Acknowledged if the RC1IE and GIE bits are set.
- 8. Read the RCSTA1 register to determine if any error occurred during reception, as well as read bit 9 of data (if applicable).
- 9. Read RCREG1 to determine if the device is being addressed.
- 10. If any error occurred, clear the CREN bit.
- 11. If the device has been addressed, clear the ADDEN bit to allow all received data into the receive buffer and interrupt the CPU.

FIGURE 19-6: EUSART RECEIVE BLOCK DIAGRAM

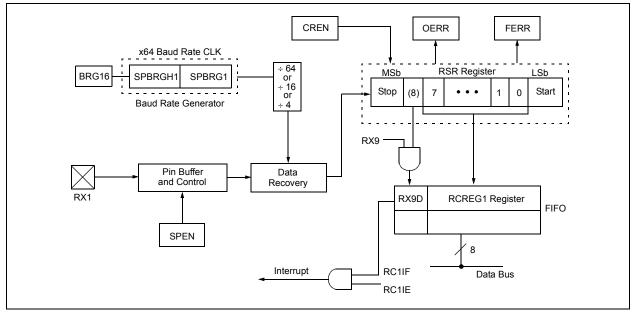


FIGURE 19-7: ASYNCHRONOUS RECEPTION

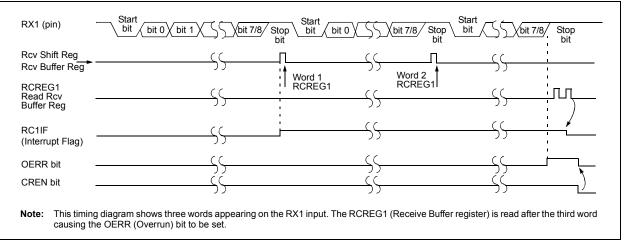


TABLE 19-6: REGISTERS ASSOCIATED WITH ASYNCHRONOUS RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page		
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59		
PIR1	_	ADIF	RC1IF	TX1IF	SSPIF	—	TMR2IF	TMR1IF	62		
PIE1	—	ADIE	RC1IE	TX1IE	SSPIE	—	TMR2IE	TMR1IE	62		
IPR1	—	ADIP	RC1IP	TX1IP	SSPIP	—	TMR2IP	TMR1IP	62		
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	61		
RCREG1	EUSART F	Receive Regi	ster						61		
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	61		
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	—	WUE	ABDEN	63		
SPBRGH1	EUSART B	USART Baud Rate Generator Register High Byte									
SPBRG1	EUSART B	aud Rate G	enerator Re	gister Low	Byte				61		

Legend: — = unimplemented locations read as '0'. Shaded cells are not used for asynchronous reception.

19.3.4 AUTO-WAKE-UP ON SYNC BREAK CHARACTER

During Sleep mode, all clocks to the EUSART are suspended. Because of this, the Baud Rate Generator is inactive and a proper byte reception cannot be performed. The auto-wake-up feature allows the controller to wake-up, due to activity on the RX1/DT1 line, while the EUSART is operating in Asynchronous mode.

The auto-wake-up feature is enabled by setting the WUE bit (BAUDCON<1>). Once set, the typical receive sequence on RX1/DT1 is disabled and the EUSART remains in an Idle state, monitoring for a wake-up event independent of the CPU mode. A wake-up event consists of a high-to-low transition on the RX1/DT1 line. (This coincides with the start of a Sync Break or a Wake-up Signal character for the LIN/J2602 protocol.)

Following a wake-up event, the module generates an RC1IF interrupt. The interrupt is generated synchronously to the Q clocks in normal operating modes (Figure 19-8), and asynchronously, if the device is in Sleep mode (Figure 19-9). The interrupt condition is cleared by reading the RCREG1 register.

The WUE bit is automatically cleared once a low-to-high transition is observed on the RX1 line following the wake-up event. At this point, the EUSART module is in Idle mode and returns to normal operation. This signals to the user that the Sync Break event is over.

19.3.4.1 Special Considerations Using Auto-Wake-up

Since auto-wake-up functions by sensing rising edge transitions on RX1/DT1, information with any state changes before the Stop bit may signal a false

End-Of-Character (EOF) and cause data or framing errors. Therefore, to work properly, the initial character in the transmission must be all '0's. This can be 00h (8 bytes) for standard RS-232 devices or 000h (12 bits) for LIN/J2602 bus.

Oscillator start-up time must also be considered, especially in applications using oscillators with longer start-up intervals (i.e., XT or HS mode). The Sync Break (or Wake-up Signal) character must be of sufficient length and be followed by a sufficient interval to allow enough time for the selected oscillator to start and provide proper initialization of the EUSART.

19.3.4.2 Special Considerations Using the WUE Bit

The timing of WUE and RC1IF events may cause some confusion when it comes to determining the validity of received data. As noted, setting the WUE bit places the EUSART in an Idle mode. The wake-up event causes a receive interrupt by setting the RC1IF bit. The WUE bit is cleared after this when a rising edge is seen on RX1/DT1. The interrupt condition is then cleared by reading the RCREG1 register. Ordinarily, the data in RCREG1 will be dummy data and should be discarded.

The fact that the WUE bit has been cleared (or is still set), and the RC1IF flag is set, should not be used as an indicator of the integrity of the data in RCREG1. Users should consider implementing a parallel method in firmware to verify received data integrity.

To assure that no actual data is lost, check the RCIDL bit to verify that a receive operation is not in process. If a receive operation is not occurring, the WUE bit may then be set just prior to entering the Sleep mode.

FIGURE 19-8: AUTO-WAKE-UP BIT (WUE) TIMINGS DURING NORMAL OPERATION

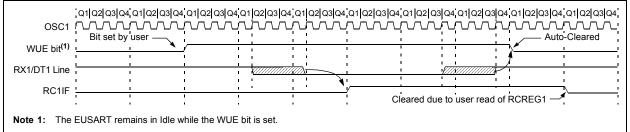
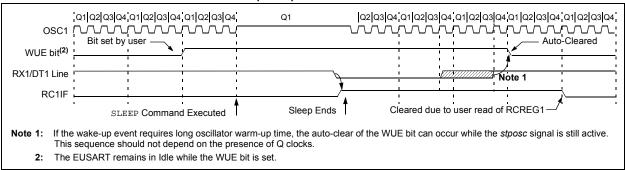


FIGURE 19-9: AUTO-WAKE-UP BIT (WUE) TIMINGS DURING SLEEP



19.3.5 BREAK CHARACTER SEQUENCE

The Enhanced USART module has the capability of sending the special Break character sequences that are required by the LIN/J2602 bus standard. The Break character transmit consists of a Start bit, followed by twelve '0' bits and a Stop bit. The Frame Break character is sent whenever the SENDB and TXEN bits (TXSTA<3> and TXSTA<5>) are set while the Transmit Shift register is loaded with data. Note that the value of data written to TXREG1 will be ignored and all '0's will be transmitted.

The SENDB bit is automatically reset by hardware after the corresponding Stop bit is sent. This allows the user to preload the transmit FIFO with the next transmit byte following the Break character (typically, the Sync character in the LIN/J2602 specification).

Note that the data value written to the TXREG1 for the Break character is ignored. The write simply serves the purpose of initiating the proper sequence.

The TRMT bit indicates when the transmit operation is active or Idle, just as it does during normal transmission. See Figure 19-10 for the timing of the Break character sequence.

19.3.5.1 Break and Sync Transmit Sequence

The following sequence will send a message frame header made up of a Break, followed by an Auto-Baud Sync byte. This sequence is typical of a LIN/J2602 bus master.

- 1. Configure the EUSART for the desired mode.
- 2. Set the TXEN and SENDB bits to set up the Break character.

- 3. Load the TXREG1 with a dummy character to initiate transmission (the value is ignored).
- 4. Write '55h' to TXREG1 to load the Sync character into the transmit FIFO buffer.
- 5. After the Break has been sent, the SENDB bit is reset by hardware. The Sync character now transmits in the preconfigured mode.

When the TXREG1 becomes empty, as indicated by the TX1IF, the next data byte can be written to TXREG1.

19.3.6 RECEIVING A BREAK CHARACTER

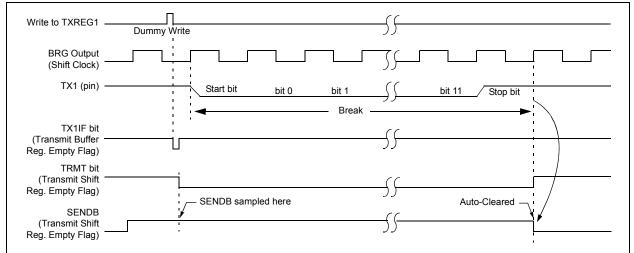
The Enhanced USART module can receive a Break character in two ways.

The first method forces configuration of the baud rate at a frequency of 9/13 the typical speed. This allows for the Stop bit transition to be at the correct sampling location (13 bits for Break versus Start bit and 8 data bits for typical data).

The second method uses the auto-wake-up feature described in **Section 19.3.4 "Auto-Wake-up On Sync Break Character"**. By enabling this feature, the EUSART will sample the next two transitions on RX1/DT1, cause an RC1IF interrupt and receive the next data byte followed by another interrupt.

Note that following a Break character, the user will typically want to enable the Auto-Baud Rate Detect feature. For both methods, the user can set the ABD bit once the TX1IF interrupt is observed.

FIGURE 19-10: SEND BREAK CHARACTER SEQUENCE



19.4 EUSART Synchronous Master Mode

The Synchronous Master mode is entered by setting the CSRC bit (TXSTA<7>). In this mode, the data is transmitted in a half-duplex manner (i.e., transmission and reception do not occur at the same time). When transmitting data, the reception is inhibited and vice versa. Synchronous mode is entered by setting bit, SYNC (TXSTA<4>). In addition, enable bit, SPEN (RCSTA1<7>), is set in order to configure the TX1 and RX1 pins to CK1 (clock) and DT1 (data) lines, respectively.

The Master mode indicates that the processor transmits the master clock on the CK1 line. Clock polarity is selected with the SCKP bit (BAUDCON<4>). Setting SCKP sets the Idle state on CK1 as high, while clearing the bit, sets the Idle state as low. This option is provided to support Microwire devices with this module.

19.4.1 EUSART SYNCHRONOUS MASTER TRANSMISSION

The EUSART transmitter block diagram is shown in Figure 19-3. The heart of the transmitter is the Transmit (Serial) Shift Register (TSR). The Shift register obtains its data from the Read/Write Transmit Buffer Register, TXREG1. The TXREG1 register is loaded with data in software. The TSR register is not loaded until the last bit has been transmitted from the previous load. As soon as the last bit is transmitted, the TSR is loaded with new data from the TXREG1 (if available). Once the TXREG1 register transfers the data to the TSR register (occurs in one TCYCLE), the TXREG1 is empty and the TX1IF flag bit (PIR1<4>) is set. The interrupt can be enabled or disabled by setting or clearing the interrupt enable bit, TX1IE (PIE1<4>). TX1IF is set regardless of the state of enable bit TX1IE; it cannot be cleared in software. It will reset only when new data is loaded into the TXREG1 register.

While flag bit, TX1IF, indicates the status of the TXREG1 register, another bit, TRMT (TXSTA<1>), shows the status of the TSR register. TRMT is a read-only bit which is set when the TSR is empty. No interrupt logic is tied to this bit so the user has to poll this bit in order to determine if the TSR register is empty. The TSR is not mapped in data memory so it is not available to the user.

To set up a Synchronous Master Transmission:

- 1. Initialize the SPBRGH1:SPBRG1 registers for the appropriate baud rate. Set or clear the BRG16 bit, as required, to achieve the desired baud rate.
- 2. Enable the synchronous master serial port by setting bits, SYNC, SPEN and CSRC.
- 3. If interrupts are desired, set enable bit, TX1IE.
- 4. If 9-bit transmission is desired, set bit, TX9.
- 5. Enable the transmission by setting bit, TXEN.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit, TX9D.
- 7. Start transmission by loading data to the TXREG1 register.
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

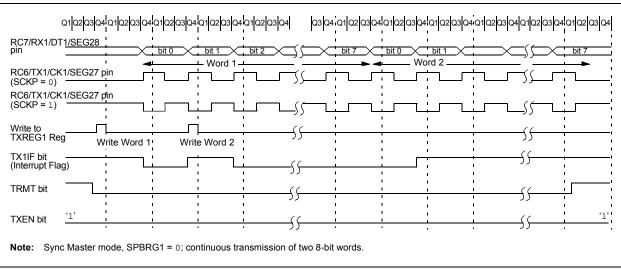


FIGURE 19-11: SYNCHRONOUS TRANSMISSION

PIC18F87J90 FAMILY

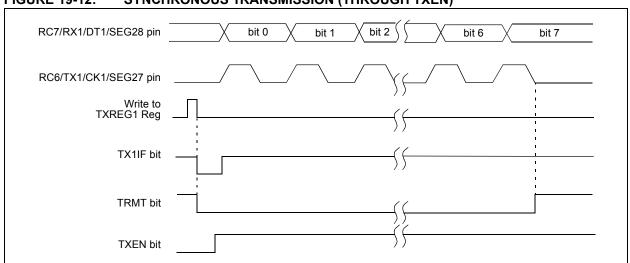


FIGURE 19-12: SYNCHRONOUS TRANSMISSION (THROUGH TXEN)

TABLE 19-7: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page	
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59	
PIR1	_	ADIF	RC1IF	TX1IF	SSPIF	_	TMR2IF	TMR1IF	62	
PIE1	_	ADIE	RC1IE	TX1IE	SSPIE	—	TMR2IE	TMR1IE	62	
IPR1	_	ADIP	RC1IP	TX1IP	SSPIP	—	TMR2IP	TMR1IP	62	
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	61	
TXREG1	EUSART T	ransmit Reg	ister						61	
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	61	
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	—	WUE	ABDEN	63	
SPBRGH1	EUSART B	EUSART Baud Rate Generator Register High Byte								
SPBRG1	EUSART B	EUSART Baud Rate Generator Register Low Byte								
LATG	U2OD	U10D		LATG4	LATG3	LATG2	LATG1	LATG0	62	

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous master transmission.

19.4.2 EUSART SYNCHRONOUS MASTER RECEPTION

Once Synchronous mode is selected, reception is enabled by setting either the Single Receive Enable bit, SREN (RCSTA1<5>), or the Continuous Receive Enable bit, CREN (RCSTA1<4>). Data is sampled on the RX1 pin on the falling edge of the clock.

If enable bit, SREN, is set, only a single word is received. If enable bit, CREN, is set, the reception is continuous until CREN is cleared. If both bits are set, then CREN takes precedence.

To set up a Synchronous Master Reception:

- 1. Initialize the SPBRGH1:SPBRG1 registers for the appropriate baud rate. Set or clear the BRG16 bit, as required, to achieve the desired baud rate.
- 2. Enable the synchronous master serial port by setting bits, SYNC, SPEN and CSRC.

- 3. Ensure bits, CREN and SREN, are clear.
- 4. If interrupts are desired, set enable bit, RC1IE.
- 5. If 9-bit reception is desired, set bit, RX9.
- 6. If a single reception is required, set bit, SREN. For continuous reception, set bit, CREN.
- 7. Interrupt flag bit, RC1IF, will be set when reception is complete and an interrupt will be generated if the enable bit, RC1IE, was set.
- 8. Read the RCSTA1 register to get the 9th bit (if enabled) and determine if any error occurred during reception.
- 9. Read the 8-bit received data by reading the RCREG1 register.
- 10. If any error occurred, clear the error by clearing bit, CREN.
- 11. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

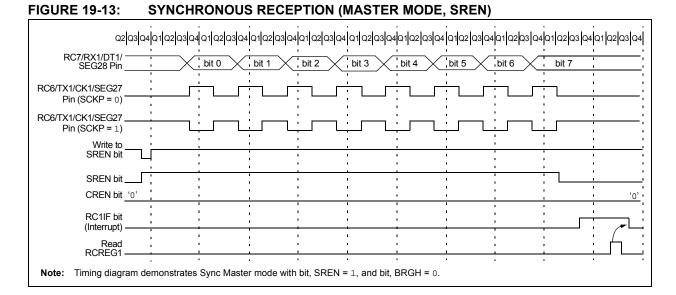


TABLE 19-8: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page		
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59		
PIR1		ADIF	RC1IF	TX1IF	SSPIF	_	TMR2IF	TMR1IF	62		
PIE1	—	ADIE	RC1IE	TX1IE	SSPIE	_	TMR2IE	TMR1IE	62		
IPR1	_	ADIP	RC1IP	TX1IP	SSPIP	—	TMR2IP	TMR1IP	62		
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	61		
RCREG1	EUSART R	eceive Regi	ster						61		
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	61		
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN	63		
SPBRGH1	EUSART B	USART Baud Rate Generator Register High Byte									
SPBRG1	EUSART Baud Rate Generator Register Low Byte										

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous master reception.

19.5 EUSART Synchronous Slave Mode

Synchronous Slave mode is entered by clearing bit, CSRC (TXSTA<7>). This mode differs from the Synchronous Master mode in that the shift clock is supplied externally at the CK1 pin (instead of being supplied internally in Master mode). This allows the device to transfer or receive data while in any Low-Power mode.

19.5.1 EUSART SYNCHRONOUS SLAVE TRANSMIT

The operation of the Synchronous Master and Slave modes is identical except in the case of Sleep mode.

If two words are written to the TXREG1 and then the SLEEP instruction is executed, the following will occur:

- a) The first word will immediately transfer to the TSR register and transmit.
- b) The second word will remain in the TXREG1 register.
- c) Flag bit, TX1IF, will not be set.
- d) When the first word has been shifted out of TSR, the TXREG1 register will transfer the second word to the TSR and flag bit, TX1IF, will now be set.
- e) If enable bit, TX1IE, is set, the interrupt will wake the chip from Sleep. If the global interrupt is enabled, the program will branch to the interrupt vector.

To set up a Synchronous Slave Transmission:

- 1. Enable the synchronous slave serial port by setting bits, SYNC and SPEN, and clearing bit, CSRC.
- 2. Clear bits, CREN and SREN.
- 3. If interrupts are desired, set enable bit, TX1IE.
- 4. If 9-bit transmission is desired, set bit, TX9.
- 5. Enable the transmission by setting enable bit, TXEN.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit, TX9D.
- 7. Start transmission by loading data to the TXREG1 register.
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page	
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59	
PIR1	—	ADIF	RC1IF	TX1IF	SSPIF	_	TMR2IF	TMR1IF	62	
PIE1	_	ADIE	RC1IE	TX1IE	SSPIE	_	TMR2IE	TMR1IE	62	
IPR1	_	ADIP	RC1IP	TX1IP	SSPIP		TMR2IP	TMR1IP	62	
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	61	
TXREG1	EUSART T	ransmit Regi	ster						61	
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	61	
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16		WUE	ABDEN	63	
SPBRGH1	EUSART B	EUSART Baud Rate Generator Register High Byte								
SPBRG1	EUSART B	USART Baud Rate Generator Register Low Byte								
LATG	U2OD	U10D		LATG4	LATG3	LATG2	LATG1	LATG0	62	

TABLE 19-9: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE TRANSMISSION

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous slave transmission.

19.5.2 EUSART SYNCHRONOUS SLAVE RECEPTION

The operation of the Synchronous Master and Slave modes is identical except in the case of Sleep or any Idle mode, and bit, SREN, which is a "don't care" in Slave mode.

If receive is enabled by setting the CREN bit prior to entering Sleep or any Idle mode, then a word may be received while in this Low-Power mode. Once the word is received, the RSR register will transfer the data to the RCREG1 register. If the RC1IE enable bit is set, the interrupt generated will wake the chip from the Low-Power mode. If the global interrupt is enabled, the program will branch to the interrupt vector. To set up a Synchronous Slave Reception:

- 1. Enable the synchronous master serial port by setting bits, SYNC and SPEN, and clearing bit, CSRC.
- 2. If interrupts are desired, set enable bit, RC1IE.
- 3. If 9-bit reception is desired, set bit, RX9.
- 4. To enable reception, set enable bit, CREN.
- 5. Flag bit, RC1IF, will be set when reception is complete. An interrupt will be generated if enable bit, RC1IE, was set.
- Read the RCSTA1 register to get the 9th bit (if enabled) and determine if any error occurred during reception.
- 7. Read the 8-bit received data by reading the RCREG1 register.
- 8. If any error occurred, clear the error by clearing bit, CREN.
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page		
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59		
PIR1	—	ADIF	RC1IF	TX1IF	SSPIF	—	TMR2IF	TMR1IF	62		
PIE1	—	ADIE	RC1IE	TX1IE	SSPIE	—	TMR2IE	TMR1IE	62		
IPR1	_	ADIP	RC1IP	TX1IP	SSPIP	_	TMR2IP	TMR1IP	62		
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	61		
RCREG1	EUSART R	Receive Regi	ster						61		
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	61		
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	—	WUE	ABDEN	63		
SPBRGH1	EUSART B	JSART Baud Rate Generator Register High Byte									
SPBRG1	EUSART B	aud Rate G	enerator Re	gister Low I	Byte				61		

TABLE 19-10: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE RECEPTION

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous slave reception.

20.0 ADDRESSABLE UNIVERSAL SYNCHRONOUS ASYNCHRONOUS RECEIVER TRANSMITTER (AUSART)

The Addressable Universal Synchronous Asynchronous Receiver Transmitter (AUSART) module is very similar in function to the Enhanced USART module, discussed in the previous chapter. It is provided as an additional channel for serial communication with external devices, for those situations that do not require auto-baud detection or LIN/J2602 bus support.

The AUSART can be configured in the following modes:

- Asynchronous (full-duplex)
- Synchronous Master (half-duplex)
- Synchronous Slave (half-duplex)

The pins of the AUSART module are multiplexed with the functions of PORTG (RG1/TX2/CK2 and RG2/RX2/DT2/VLCAP1, respectively). In order to configure these pins as an AUSART:

- bit, SPEN (RCSTA2<7>), must be set (= 1)
- bit, TRISG<2>, must be set (= 1)
- bit, TRISG<1>, must be cleared (= 0) for Asynchronous and Synchronous Master modes
- bit, TRISG<1>, must be set (= 1) for Synchronous Slave mode

Note: The AUSART control will automatically reconfigure the pin from input to output as needed.

The driver for the TX2 output pin can also be optionally configured as an open-drain output. This feature allows the voltage level on the pin to be pulled to a higher level through an external pull-up resistor, and allows the output to communicate with external circuits without the need for additional level shifters.

The open-drain output option is controlled by the U2OD bit (LATG<7>). Setting the bit configures the pin for open-drain operation.

20.1 Control Registers

The operation of the Addressable USART module is controlled through two registers, TXSTA2 and RXSTA2. These are detailed in Register 20-1 and Register 20-2, respectively.

PIC18F87J90 FAMILY

REGISTER				STATUS AN			
R/W-0	R/W-0	R/W-0 TXEN ⁽¹⁾	R/W-0	U-0	R/W-0	R-1	R/W-0
CSRC	TX9	I XEIN''	SYNC		BRGH	TRMT	TX9D bit
_egend:							
R = Readable	e bit	W = Writable	bit	U = Unimpler	nented bit, rea	ad as '0'	
n = Value at	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unki	nown
pit 7	CSRC: Clock	Source Select	bit				
	Asynchronous Don't care.	<u>s mode:</u>					
		<u>mode:</u> ode (clock gene de (clock from e					
pit 6	TX9: 9-Bit Tra 1 = Selects 9	ansmit Enable k -bit transmissio -bit transmissio	pit on				
oit 5		mit Enable bit ⁽¹					
	0 = Transmit	disabled					
oit 4		RT Mode Sele	ct bit				
	1 = Synchror 0 = Asynchro						
oit 3	-	ted: Read as ')'				
oit 2	-	Baud Rate Sele					
	Asynchronous						
	1 = High spend 0 = Low spend						
	Synchronous						
	Unused in this						
oit 1	TRMT: Transr	mit Shift Regist	er Status bit				
	1 = TSR emp 0 = TSR full	oty					
oit 0	TX9D: 9th bit	of Transmit Da	ta				
	Can be addre	ee/data hit or a	parity bit				
			parity bit.				

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-0	R-0	R-x
SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D
oit 7							bit
Legend:							
R = Readab	le bit	W = Writable	oit	U = Unimplem	ented bit, read	d as '0'	
-n = Value a	t POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	iown
bit 7	SPEN: Seria	I Port Enable bit	:				
		ort enabled (con ort disabled (hel		T2 and TX2/CK	2 (TXEN = 1)	pins as serial p	ort pins)
bit 6	RX9: 9-Bit R	eceive Enable b	it				
		9-bit reception 8-bit reception					
oit 5	SREN: Sing	le Receive Enab	le bit				
	<u>Asynchrono</u> Don't care.	<u>us mode</u> :					
	1 = Enables 0 = Disables	s mode – Master s single receive s single receive eared after recep		ete.			
		s mode – Slave:	•				
bit 4	CREN: Cont	inuous Receive	Enable bit				
	Asynchronou 1 = Enables 0 = Disables	receiver					
		<u>s mode:</u> continuous rece s continuous rec		ole bit, CREN, is	cleared (CRE	N overrides SR	EN)
bit 3	Asynchronou 1 = Enables 0 = Disables	dress Detect En us mode 9-bit (R address detecti s address detect us mode 9-bit (R	<u>X9 = 1)</u> : on, enables ir ion, all bytes a				
bit 2	FERR: Fram 1 = Framing 0 = No fram	error (can be cl	eared by reac	ling the RCREG	x register and	receiving the n	ext valid byte
oit 1	OERR: Ove	rrun Error bit error (can be cl	eared by clea	ring bit CREN)			
oit 0		it of Received D	ata				
лго			ulu				

20.2 AUSART Baud Rate Generator (BRG)

The BRG is a dedicated, 8-bit generator that supports both the Asynchronous and Synchronous modes of the AUSART.

The SPBRG2 register controls the period of a free-running timer. In Asynchronous mode, the BRGH bit (TXSTA<2>) also controls the baud rate. In Synchronous mode, BRGH is ignored. Table 20-1 shows the formula for computation of the baud rate for different AUSART modes, which only apply in Master mode (internally generated clock).

Given the desired baud rate and FOSC, the nearest integer value for the SPBRG2 register can be calculated using the formulas in Table 20-1. From this, the error in baud rate can be determined. An example calculation is shown in Example 20-1. Typical baud rates and error values for the various Asynchronous modes are shown in Table 20-2. It may be advantageous to use the high baud rate (BRGH = 1) to reduce the baud rate error, or achieve a slow baud rate for a fast oscillator frequency.

Writing a new value to the SPBRG2 register causes the BRG timer to be reset (or cleared). This ensures the BRG does not wait for a timer overflow before outputting the new baud rate.

20.2.1 OPERATION IN POWER-MANAGED MODES

The device clock is used to generate the desired baud rate. When one of the power-managed modes is entered, the new clock source may be operating at a different frequency. This may require an adjustment to the value in the SPBRG2 register.

20.2.2 SAMPLING

The data on the RX2 pin is sampled three times by a majority detect circuit to determine if a high or a low level is present at the RX2 pin.

TABLE 20-1: BAUD RATE FORMULAS

Configura	iguration Bits BRGH 0 1		David Data Farmula
SYNC	BRGH	BRG/AUSART Mode	Baud Rate Formula
0	0	Asynchronous	Fosc/[64 (n + 1)]
0	1	Asynchronous	Fosc/[16 (n + 1)]
1	x	Synchronous	Fosc/[4 (n + 1)]

Legend: x = Don't care, n = Value of SPBRG2 register

EXAMPLE 20-1: CALCULATING BAUD RATE ERROR

For a device with Fost	c of	16 MHz, desired baud rate of 9600, Asynchronous mode, BRGH = 0:
Desired Baud Rate	=	Fosc/(64 ([SPBRG2] + 1))
Solving for SPBRG2:		
Х	=	((FOSC/Desired Baud Rate)/64) - 1
	=	((1600000/9600)/64) – 1
	=	[25.042] = 25
Calculated Baud Rate	=	1600000/(64 (25 + 1))
	=	9615
Error	=	(Calculated Baud Rate - Desired Baud Rate)/Desired Baud Rate
	=	(9615 - 9600)/9600 = 0.16%

TABLE 20-2: REGISTERS ASSOCIATED WITH THE BAUD RATE GENERATOR

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page	
TXSTA2	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	64	
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	64	
SPBRG2	AUSART E	USART Baud Rate Generator Register								

Legend: Shaded cells are not used by the BRG.

						BRG	H = 0					
	Fosc	= 40.000) MHz	Fosc	= 20.000) MHz	Foso	: = 10.000	MHz	Fosc = 8.000 MHz		
BAUD RATE (K)	Actual Rate (K)	% Error	SPBRG value (decimal)									
0.3				_			_		_	_		_
1.2	—	—	_	1.221	1.73	255	1.202	0.16	129	1.201	-0.16	103
2.4	2.441	1.73	255	2.404	0.16	129	2.404	0.16	64	2.403	-0.16	51
9.6	9.615	0.16	64	9.766	1.73	31	9.766	1.73	15	9.615	-0.16	12
19.2	19.531	1.73	31	19.531	1.73	15	19.531	1.73	7	_	_	_
57.6	56.818	-1.36	10	62.500	8.51	4	52.083	-9.58	2	—	_	_
115.2	125.000	8.51	4	104.167	-9.58	2	78.125	-32.18	1	_	_	—

TABLE 20-3: BAUD RATES FOR ASYNCHRONOUS MODES

					BRGH =	0				
	Fos	c = 4.000	MHz	Fos	c = 2.000	MHz	Fosc = 1.000 MHz			
BAUD RATE (K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	
0.3	0.300	0.16	207	0.300	-0.16	103	0.300	-0.16	51	
1.2	1.202	0.16	51	1.201	-0.16	25	1.201	-0.16	12	
2.4	2.404	0.16	25	2.403	-0.16	12	_	_	_	
9.6	8.929	-6.99	6	—	_	_	_	_	_	
19.2	20.833	8.51	2	—	_	_	_	_	_	
57.6	62.500	8.51	0	—	_	_	—	_	_	
115.2	62.500	-45.75	0	_	—	—	_	—	—	

		BRGH = 1														
BAUD RATE	Fosc	= 40.000) MHz	Fosc = 20.000 MHz			Fosc = 10.000 MHz			Fos	Fosc = 8.000 MHz					
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)				
0.3										_		_				
1.2	—	_	_	—	_	_	_	_	—	_	_	—				
2.4	—	_	_	—	_	_	2.441	1.73	255	2.403	-0.16	207				
9.6	9.766	1.73	255	9.615	0.16	129	9.615	0.16	64	9.615	-0.16	51				
19.2	19.231	0.16	129	19.231	0.16	64	19.531	1.73	31	19.230	-0.16	25				
57.6	58.140	0.94	42	56.818	-1.36	21	56.818	-1.36	10	55.555	3.55	8				
115.2	113.636	-1.36	21	113.636	-1.36	10	125.000	8.51	4	—	_	_				

					BRGH =	1				
BAUD	Foso	; = 4.000	MHz	Fos	c = 2.000	MHz	Fosc = 1.000 MHz			
(K)	Actual Rate (K)	e [%] value		Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	
0.3				_		_	0.300	-0.16	207	
1.2	1.202	0.16	207	1.201	-0.16	103	1.201	-0.16	51	
2.4	2.404	0.16	103	2.403	-0.16	51	2.403	-0.16	25	
9.6	9.615	0.16	25	9.615	-0.16	12	_	_	_	
19.2	19.231	0.16	12	_	_	_	_	_	_	
57.6	62.500	8.51	3	—	_	_	—	_	_	
115.2	125.000	8.51	1	_	—	—	_			

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20.3 AUSART Asynchronous Mode

The Asynchronous mode of operation is selected by clearing the SYNC bit (TXSTA2<4>). In this mode, the AUSART uses standard Non-Return-to-Zero (NRZ) format (one Start bit, eight or nine data bits and one Stop bit). The most common data format is 8 bits. An on-chip, dedicated, 8-bit Baud Rate Generator can be used to derive standard baud rate frequencies from the oscillator.

The AUSART transmits and receives the LSb first. The AUSART's transmitter and receiver are functionally independent but use the same data format and baud rate. The Baud Rate Generator produces a clock, either x16 or x64 of the bit shift rate, depending on the BRGH bit (TXSTA2<2>). Parity is not supported by the hardware but can be implemented in software and stored as the 9th data bit.

When operating in Asynchronous mode, the AUSART module consists of the following important elements:

- · Baud Rate Generator
- · Sampling Circuit
- Asynchronous Transmitter
- · Asynchronous Receiver

20.3.1 AUSART ASYNCHRONOUS TRANSMITTER

The AUSART transmitter block diagram is shown in Figure 20-1. The heart of the transmitter is the Transmit (Serial) Shift Register (TSR). The Shift register obtains its data from the Read/Write Transmit Buffer Register, TXREG2. The TXREG2 register is loaded with data in software. The TSR register is not loaded until the Stop bit has been transmitted from the previous load. As soon as the Stop bit is transmitted, the TSR is loaded with new data from the TXREG2 register (if available). Once the TXREG2 register transfers the data to the TSR register (occurs in one TcY), the TXREG2 register is empty and the TX2IF flag bit (PIR3<4>) is set. This interrupt can be enabled or disabled by setting or clearing the interrupt enable bit, TX2IE (PIE3<4>). TX2IF will be set regardless of the state of TX2IE; it cannot be cleared in software. TX2IF is also not cleared immediately upon loading TXREG2, but becomes valid in the second instruction cycle following the load instruction. Polling TX2IF immediately following a load of TXREG2 will return invalid results.

While TX2IF indicates the status of the TXREG2 register, another bit, TRMT (TXSTA2<1>), shows the status of the TSR register. TRMT is a read-only bit which is set when the TSR register is empty. No interrupt logic is tied to this bit so the user has to poll this bit in order to determine if the TSR register is empty.

Note 1:	The TSR register is not mapped in data							
	memory so it is not available to the user.							
2:	Flag bit, TX2IF, is set when enable bit, TXEN, is set.							

To set up an Asynchronous Transmission:

- 1. Initialize the SPBRG2 register for the appropriate baud rate. Set or clear the BRGH bit, as required, to achieve the desired baud rate.
- 2. Enable the asynchronous serial port by clearing bit, SYNC, and setting bit, SPEN.
- 3. If interrupts are desired, set enable bit, TX2IE.
- 4. If 9-bit transmission is desired, set transmit bit, TX9. Can be used as address/data bit.
- 5. Enable the transmission by setting bit, TXEN, which will also set bit, TX2IF.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit, TX9D.
- 7. Load data to the TXREG2 register (starts transmission).
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

Data Bus TX2IF TXREG2 Register TX2IE 8 MSb LSb Pin Buffer (8) 0 . . . and Control TSR Register TX2 pin Interrupt Baud Rate CLK TXFN SPEN TRMT SPBRG2 тхс Baud Rate Generator TX9D

FIGURE 20-1: AUSART TRANSMIT BLOCK DIAGRAM

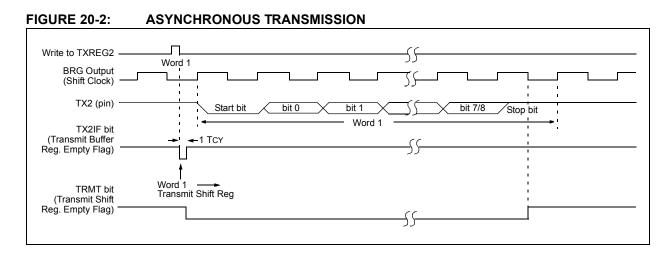


FIGURE 20-3: ASYNCHRONOUS TRANSMISSION (BACK TO BACK)

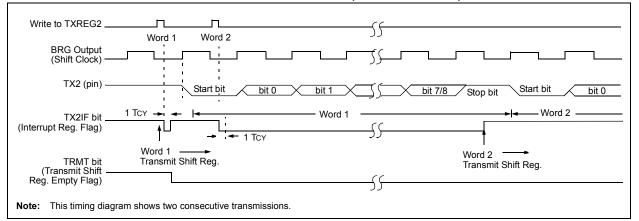


TABLE 20-4: REGISTERS ASSOCIATED WITH ASYNCHRONOUS TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59
PIR3	—	LCDIF	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF	62
PIE3	—	LCDIE	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE	62
IPR3	—	LCDIP	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP	62
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	64
TXREG2	AUSART T	ransmit Reg	ister						64
TXSTA2	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	64
SPBRG2	AUSART B	aud Rate G	enerator Re	gister					64
LATG	U2OD	U10D	_	LATG4	LATG3	LATG2	LATG1	LATG0	62

Legend: — = unimplemented locations read as '0'. Shaded cells are not used for asynchronous transmission.

20.3.2 AUSART ASYNCHRONOUS RECEIVER

The receiver block diagram is shown in Figure 20-4. The data is received on the RX2 pin and drives the data recovery block. The data recovery block is actually a high-speed shifter operating at x16 times the baud rate, whereas the main receive serial shifter operates at the bit rate or at Fosc. This mode would typically be used in RS-232 systems.

To set up an Asynchronous Reception:

- 1. Initialize the SPBRG2 register for the appropriate baud rate. Set or clear the BRGH bit, as required, to achieve the desired baud rate.
- 2. Enable the asynchronous serial port by clearing bit, SYNC, and setting bit, SPEN.
- 3. If interrupts are desired, set enable bit, RC2IE.
- 4. If 9-bit reception is desired, set bit, RX9.
- 5. Enable the reception by setting bit, CREN.
- Flag bit, RC2IF, will be set when reception is complete and an interrupt will be generated if enable bit, RC2IE, was set.
- 7. Read the RCSTA2 register to get the 9th bit (if enabled) and determine if any error occurred during reception.
- 8. Read the 8-bit received data by reading the RCREG2 register.
- 9. If any error occurred, clear the error by clearing enable bit, CREN.
- 10. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

20.3.3 SETTING UP 9-BIT MODE WITH ADDRESS DETECT

This mode would typically be used in RS-485 systems. To set up an Asynchronous Reception with Address Detect Enable:

- 1. Initialize the SPBRG2 register for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
- 2. Enable the asynchronous serial port by clearing the SYNC bit and setting the SPEN bit.
- If interrupts are required, set the RCEN bit and select the desired priority level with the RC2IP bit.
- 4. Set the RX9 bit to enable 9-bit reception.
- 5. Set the ADDEN bit to enable address detect.
- 6. Enable reception by setting the CREN bit.
- The RC2IF bit will be set when reception is complete. The interrupt will be Acknowledged if the RC2IE and GIE bits are set.
- 8. Read the RCSTA2 register to determine if any error occurred during reception, as well as read bit 9 of data (if applicable).
- 9. Read RCREG2 to determine if the device is being addressed.
- 10. If any error occurred, clear the CREN bit.
- 11. If the device has been addressed, clear the ADDEN bit to allow all received data into the receive buffer and interrupt the CPU.

FIGURE 20-4: AUSART RECEIVE BLOCK DIAGRAM

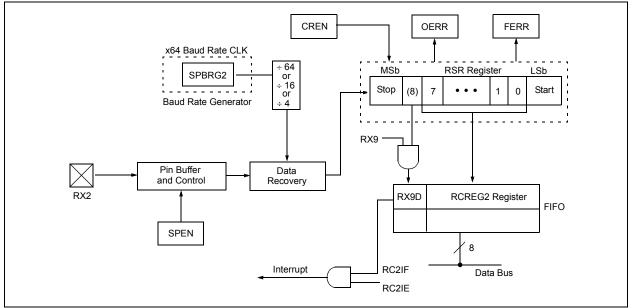


FIGURE 20-5: ASYNCHRONOUS RECEPTION

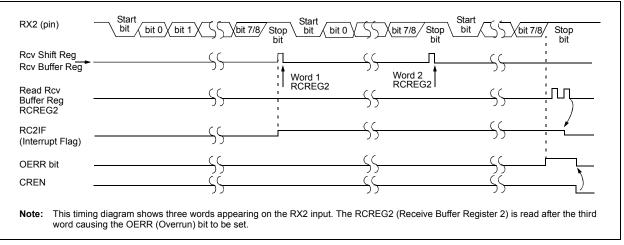


TABLE 20-5: REGISTERS ASSOCIATED WITH ASYNCHRONOUS RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page	
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59	
PIR3	—	LCDIF	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF	62	
PIE3	—	LCDIE	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE	62	
IPR3		LCDIP	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP	62	
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	64	
RCREG2	AUSART Receive Register									
TXSTA2	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	64	
SPBRG2	AUSART Baud Rate Generator Register									

Legend: — = unimplemented locations read as '0'. Shaded cells are not used for asynchronous reception.

20.4 AUSART Synchronous Master Mode

The Synchronous Master mode is entered by setting the CSRC bit (TXSTA2<7>). In this mode, the data is transmitted in a half-duplex manner (i.e., transmission and reception do not occur at the same time). When transmitting data, the reception is inhibited and vice versa. Synchronous mode is entered by setting bit, SYNC (TXSTA2<4>). In addition, enable bit, SPEN (RCSTA2<7>), is set in order to configure the TX2 and RX2 pins to CK2 (clock) and DT2 (data) lines, respectively.

The Master mode indicates that the processor transmits the master clock on the CK2 line.

20.4.1 AUSART SYNCHRONOUS MASTER TRANSMISSION

The AUSART transmitter block diagram is shown in Figure 20-1. The heart of the transmitter is the Transmit (Serial) Shift Register (TSR). The Shift register obtains its data from the Read/Write Transmit Buffer Register, TXREG2. The TXREG2 register is loaded with data in software. The TSR register is not loaded until the last bit has been transmitted from the previous load. As soon as the last bit is transmitted, the TSR is loaded with new data from the TXREG2 (if available). Once the TXREG2 register transfers the data to the TSR register (occurs in one TCYCLE), the TXREG2 is empty and the TX2IF flag bit (PIR3<4>) is set. The interrupt can be enabled or disabled by setting or clearing the interrupt enable bit, TX2IE (PIE3<4>). TX2IF is set regardless of the state of enable bit, TX2IE; it cannot be cleared in software. It will reset only when new data is loaded into the TXREG2 register.

While flag bit, TX2IF, indicates the status of the TXREG2 register, another bit, TRMT (TXSTA2<1>), shows the status of the TSR register. TRMT is a read-only bit which is set when the TSR is empty. No interrupt logic is tied to this bit so the user has to poll this bit in order to determine if the TSR register is empty. The TSR is not mapped in data memory so it is not available to the user.

To set up a Synchronous Master Transmission:

- 1. Initialize the SPBRG2 register for the appropriate baud rate.
- 2. Enable the synchronous master serial port by setting bits, SYNC, SPEN and CSRC.
- 3. If interrupts are desired, set enable bit, TX2IE.
- 4. If 9-bit transmission is desired, set bit, TX9.
- 5. Enable the transmission by setting bit, TXEN.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit, TX9D.
- 7. Start transmission by loading data to the TXREG2 register.
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

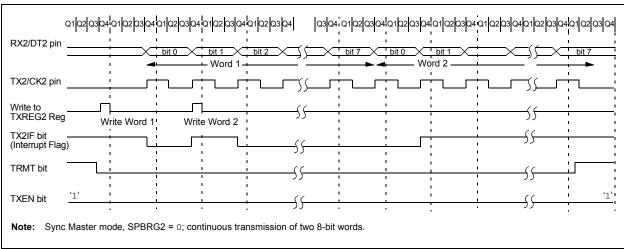


FIGURE 20-6: SYNCHRONOUS TRANSMISSION

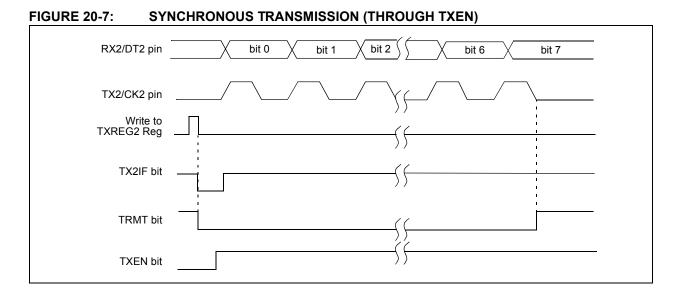


TABLE 20-6: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59
PIR3		LCDIF	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF	62
PIE3	_	LCDIE	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE	62
IPR3	_	LCDIP	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP	62
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	64
TXREG2	AUSART Transmit Register								
TXSTA2	CSRC	TX9	TXEN	SYNC	_	BRGH	TRMT	TX9D	64
SPBRG2	AUSART Baud Rate Generator Register								
LATG	U2OD	U10D	_	LATG4	LATG3	LATG2	LATG1	LATG0	62

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous master transmission.

20.4.2 AUSART SYNCHRONOUS MASTER RECEPTION

Once Synchronous mode is selected, reception is enabled by setting either the Single Receive Enable bit, SREN (RCSTA2<5>), or the Continuous Receive Enable bit, CREN (RCSTA2<4>). Data is sampled on the RX2 pin on the falling edge of the clock.

If enable bit, SREN, is set, only a single word is received. If enable bit, CREN, is set, the reception is continuous until CREN is cleared. If both bits are set, then CREN takes precedence.

To set up a Synchronous Master Reception:

- 1. Initialize the SPBRG2 register for the appropriate baud rate.
- 2. Enable the synchronous master serial port by setting bits, SYNC, SPEN and CSRC.
- 3. Ensure bits, CREN and SREN, are clear.

- 4. If interrupts are desired, set enable bit, RC2IE.
- 5. If 9-bit reception is desired, set bit, RX9.
- 6. If a single reception is required, set bit, SREN. For continuous reception, set bit, CREN.
- 7. Interrupt flag bit, RC2IF, will be set when reception is complete and an interrupt will be generated if the enable bit, RC2IE, was set.
- 8. Read the RCSTA2 register to get the 9th bit (if enabled) and determine if any error occurred during reception.
- 9. Read the 8-bit received data by reading the RCREG2 register.
- 10. If any error occurred, clear the error by clearing bit, CREN.
- 11. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

FIGURE 20-8: SYNCHRONOUS RECEPTION (MASTER MODE, SREN)

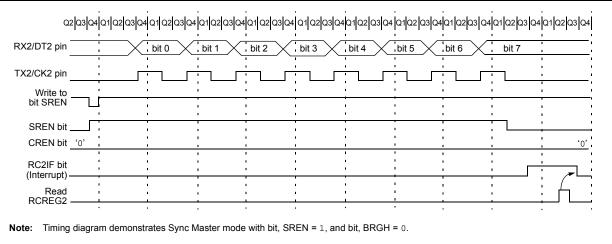


TABLE 20-7: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page					
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59					
PIR3	—	LCDIF	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF	62					
PIE3	—	LCDIE	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE	62					
IPR3	—	LCDIP	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP	62					
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	64					
RCREG2	AUSART Receive Register													
TXSTA2	CSRC	TX9	TXEN	SYNC	_	BRGH	TRMT	TX9D	64					
SPBRG2	AUSART B	aud Rate Ge	enerator Reg	gister	AUSART Baud Rate Generator Register									

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous master reception.

20.5 AUSART Synchronous Slave Mode

Synchronous Slave mode is entered by clearing bit, CSRC (TXSTA2<7>). This mode differs from the Synchronous Master mode in that the shift clock is supplied externally at the CK2 pin (instead of being supplied internally in Master mode). This allows the device to transfer or receive data while in any Low-Power mode.

20.5.1 AUSART SYNCHRONOUS SLAVE TRANSMIT

The operation of the Synchronous Master and Slave modes are identical except in the case of the Sleep mode.

If two words are written to the TXREG2 and then the SLEEP instruction is executed, the following will occur:

- a) The first word will immediately transfer to the TSR register and transmit.
- b) The second word will remain in the TXREG2 register.
- c) Flag bit, TX2IF, will not be set.
- d) When the first word has been shifted out of TSR, the TXREG2 register will transfer the second word to the TSR and flag bit, TX2IF, will now be set.
- e) If enable bit, TX2IE, is set, the interrupt will wake the chip from Sleep. If the global interrupt is enabled, the program will branch to the interrupt vector.

To set up a Synchronous Slave Transmission:

- Enable the synchronous slave serial port by setting bits, SYNC and SPEN, and clearing bit, CSRC.
- 2. Clear bits, CREN and SREN.
- 3. If interrupts are desired, set enable bit, TX2IE.
- 4. If 9-bit transmission is desired, set bit, TX9.
- 5. Enable the transmission by setting enable bit, TXEN.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit, TX9D.
- 7. Start transmission by loading data to the TXREG2 register.
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page	
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59	
PIR3	—	LCDIF	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF	62	
PIE3	—	LCDIE	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE	62	
IPR3	—	LCDIP	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP	62	
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	64	
TXREG2	AUSART Transmit Register									
TXSTA2	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	64	
SPBRG2	AUSART Baud Rate Generator Register									
LATG	U2OD	U10D		LATG4	LATG3	LATG2	LATG1	LATG0	62	

TABLE 20-8: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE TRANSMISSION

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous slave transmission.

20.5.2 AUSART SYNCHRONOUS SLAVE RECEPTION

The operation of the Synchronous Master and Slave modes is identical, except in the case of Sleep or any Idle mode, and bit, SREN, which is a "don't care" in Slave mode.

If receive is enabled by setting the CREN bit prior to entering Sleep, or any Idle mode, then a word may be received while in this Low-Power mode. Once the word is received, the RSR register will transfer the data to the RCREG2 register; if the RC2IE enable bit is set, the interrupt generated will wake the chip from Low-Power mode. If the global interrupt is enabled, the program will branch to the interrupt vector.

To set up a Synchronous Slave Reception:

- Enable the synchronous master serial port by 1. setting bits, SYNC and SPEN, and clearing bit, CSRC.
- 2. If interrupts are desired, set enable bit, RC2IE.
- 3. If 9-bit reception is desired, set bit, RX9.
- 4. To enable reception, set enable bit, CREN.
- 5. Flag bit, RC2IF, will be set when reception is complete. An interrupt will be generated if enable bit, RC2IE, was set.
- 6. Read the RCSTA2 register to get the 9th bit (if enabled) and determine if any error occurred during reception.
- 7. Read the 8-bit received data by reading the RCREG2 register.
- 8. If any error occurred, clear the error by clearing bit, CREN.
- 9. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

TABLE 20-9	: REGIS	REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE RECEPTION									
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	,		

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59
PIR3	—	LCDIF	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF	62
PIE3	—	LCDIE	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE	62
IPR3	—	LCDIP	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP	62
RCSTA2	SPEN RX9 SREN CREN ADDEN FERR OERR RX9D								64
RCREG2	AUSART Receive Register								
TXSTA2	CSRC	TX9	TXEN	SYNC	_	BRGH	TRMT	TX9D	64
SPBRG2	AUSART B	aud Rate Ge	enerator Re	gister					64

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous slave reception.

The ADCON0 register, shown in Register 21-1,

controls the operation of the A/D module. The

ADCON1 register, shown in Register 21-2, configures

21.0 10-BIT ANALOG-TO-DIGITAL CONVERTER (A/D) MODULE

The Analog-to-Digital (A/D) Converter module has 12 inputs for all PIC18F87J90 family devices. This module allows conversion of an analog input signal to a corresponding 10-bit digital number.

The module has five registers:

- A/D Result High Register (ADRESH)
- A/D Result Low Register (ADRESL)
- A/D Control Register 0 (ADCON0)
- A/D Control Register 1 (ADCON1)
- A/D Control Register 2 (ADCON2)

REGISTER 21-1: ADCON0: A/D CONTROL REGISTER 0

R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
ADCAL	—	CHS3	CHS2	CHS1	CHS0	GO/DONE	ADON
bit 7							bit 0

Legend:											
R = Readable bit		W = Writable bit	U = Unimplemented bit	, read as '0'							
-n = Value	at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown							
bit 7		A/D Calibration bit									
		pration is performed on next A nal A/D Converter operation (
bit 6	Unimple	mented: Read as '0'									
bit 5-2	CHS<3:0	>: Analog Channel Select bi	ts								
	0000 = 0	Channel 00 (AN0)									
	0001 = 0	Channel 01 (AN1)									
	0010 = (0010 = Channel 02 (AN2)									
	0011 = Channel 03 (AN3)										
	0100 = Channel 04 (AN4)										
	0101 = Channel 05 (AN5)										
		0110 = Channel 06 (AN6)									
		0111 = Channel 07 (AN7)									
	1000 = Channel 08 (AN8) 1001 = Channel 09 (AN9)										
	1010 = Channel 10 (AN10)										
	1011 = Channel 11 (AN11)										
	11xx = Unused										
bit 1	GO/DON	E: A/D Conversion Status bit	t								
	When ADON = 1:										
	1 = A/D conversion in progress										
	0 = A/D	Idle									
bit 0	ADON: A	VD On bit									
		Converter module is enabled									
	0 = A/D	Converter module is disabled									

the functions of the port pins. The ADCON2 register, shown in Register 21-3, configures the A/D clock source, programmed acquisition time and justification.

	U-0	R/	W-0	R/	W-0	R/	W-0	R	/W-0	R	2/W-0	R	R/W-0
TRIGSEL	—	VC	FG1	VC	FG0	PC	FG3	PC	CFG2	P	CFG1	P	CFG0
bit 7				•									bit
Legend:													
R = Readabl			Vritable				-		l bit, rea	ad as '0			
-n = Value at	POR	'1' = E	Bit is set			'O' = E	Bit is cle	eared		x = B	Bit is unl	known	
bit 7	TRIGSEL: Sp		riggor S	oloot bi	+								
JIL 7						MU							
	 1 = Selects the special trigger from the CTMU 0 = Selects the special trigger from the CCP2 												
oit 6	Unimplemen	ted: Re	ead as '	0'									
oit 5	VCFG1: Volta		erence	Configu	uration I	oit (VRE	F- sour	ce)					
	1 = VREF - (A)	N2)											
pit 4	0 = AVss VCFG0: Volta		aranco	Config	Ination								
JIL 4	1 = VREF+ (A	•	erence	Connigl	auon	JIL (VRE	rt soul	Ce)					
	0 = AVDD												
oit 3-0	PCFG<3:0>: A/D Port Configuration Control bits:												
	PCFG<3:0>	AN11	AN10	AN9	AN8	AN7	AN6	AN5	AN4	AN3	AN2	AN1	AN0
	FCFG-3.02	ANTI	ANIU	ANS	ANO		AND		AIN4	ANJ	ANZ		
	0000	۸	^	^	۸	^				۸	۸	^	
	0000	A	A	A	A	A	A	Α	Α	A	A	A	A
	0001	Α	Α	А	Α	А	A A	A A	A A	Α	Α	Α	A A
	0001	A A	A A	A A	A A	A A	A A A	A A A	A A A	A A	A A	A A	A A A
	0001 0010 0011	A A A	A A A	A A A	A A A	A A A	A A A A	A A A A	A A A A	A A A	A A A	A A A	A A A A
	0001	A A A D	A A A A	A A A A	A A	A A	A A A	A A A A A	A A A	A A A A	A A	A A A A	A A A A A
	0001 0010 0011	A A A	A A A	A A A	A A A	A A A	A A A A	A A A A	A A A A	A A A	A A A	A A A	A A A A
	0001 0010 0011 0100	A A A D	A A A A	A A A A	A A A A	A A A A	A A A A A	A A A A A	A A A A A	A A A A	A A A A	A A A A	A A A A A
	0001 0010 0011 0100 0101	A A A D D	A A A D	A A A A A	A A A A A	A A A A A	A A A A A A	A A A A A A	A A A A A A	A A A A A	A A A A A	A A A A A	A A A A A A
	0001 0010 0011 0100 0101 0110	A A D D D	A A A D D	A A A A D	A A A A A A	A A A A A A	A A A A A A A	A A A A A A A	A A A A A A A	A A A A A A	A A A A A A	A A A A A A	A A A A A A A
	0001 0010 0011 0100 0101 0110 0111	A A D D D D D	A A A D D D	A A A A D D	A A A A A A D	A A A A A A A	A A A A A A A A	A A A A A A A A	A A A A A A A A	A A A A A A A	A A A A A A A	A A A A A A A	A A A A A A A A
	0001 0010 0011 0100 0101 0110 0111 1000	A A D D D D D D	A A A D D D D D	A A A A D D D D	A A A A A D D	A A A A A A A D	A A A A A A A A A	A A A A A A A A A	A A A A A A A A A	A A A A A A A A A	A A A A A A A A A	A A A A A A A A A	A A A A A A A A
	0001 0010 0011 0100 0101 0110 0111 1000 1001	A A D D D D D D D D D	A A A D D D D D D D	A A A A D D D D D	A A A A A A D D D D	A A A A A A D D	A A A A A A A A A A D	A A A A A A A A A A	A A A A A A A A A A A	A A A A A A A A A	A A A A A A A A A	A A A A A A A A A	A A A A A A A A A A
	0001 0010 0011 0100 0101 0110 0111 1000 1001 1010	A A D D D D D D D D D D	A A A D D D D D D D D D	A A A A D D D D D D D D	A A A A A D D D D D	A A A A A A A D D D D	A A A A A A A A A D D D	A A A A A A A A A D	A A A A A A A A A A A	A A A A A A A A A A A	A A A A A A A A A A A	A A A A A A A A A A A	A A A A A A A A A A A
	0001 0010 0011 0100 0101 0110 0111 1000 1001 1010 1011	A A D D D D D D D D D D D D D	A A A D D D D D D D D D D D	A A A A D D D D D D D D D D	A A A A A D D D D D D D D	A A A A A A A D D D D D D	A A A A A A A A A D D D D	A A A A A A A A A A D D	A A A A A A A A A A D	A A A A A A A A A A A	A A A A A A A A A A A	A A A A A A A A A A A	A A A A A A A A A A A
	0001 0010 0011 0100 0101 0110 0111 1000 1001 1010 1011 1100	A A D D D D D D D D D D D D D	A A A D D D D D D D D D D D D	A A A A D D D D D D D D D D D	A A A A A D D D D D D D D D	A A A A A A D D D D D D D	A A A A A A A A A D D D D D D	A A A A A A A A A D D D D	A A A A A A A A A A D D	A A A A A A A A A A D	A A A A A A A A A A A A A	A A A A A A A A A A A A	A A A A A A A A A A A A A

REGISTER 21-2: ADCON1: A/D CONTROL REGISTER 1

A = Analog input

D = Digital I/O

R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
ADFM	—	ACQT2	ACQT1	ACQT0	ADCS2	ADCS1	ADCS0
bit 7							bit (
Legend:							
R = Readal	ble bit	W = Writable I	oit	U = Unimplem	nented bit, rea	d as '0'	
-n = Value a	at POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown
bit 7	ADFM: A/D R 1 = Right justi 0 = Left justifie		elect bit				
bit 6	Unimplement	ted: Read as 'o)'				
	$111 = 20 \text{ TAD} \\ 110 = 16 \text{ TAD} \\ 101 = 12 \text{ TAD} \\ 100 = 8 \text{ TAD} \\ 011 = 6 \text{ TAD} \\ 010 = 4 \text{ TAD} \\ 001 = 2 \text{ TAD} \\ 000 = 0 \text{ TAD}^{(1)}$)					
bit 2-0	111 = FRC (cl 110 = Fosc/6 101 = Fosc/1 100 = Fosc/4	6 ock derived fro 2	m A/D RC osc	cillator) ⁽¹⁾			

REGISTER 21-3: ADCON2: A/D CONTROL REGISTER 2

Note 1: If the A/D FRC clock source is selected, a delay of one TcY (instruction cycle) is added before the A/D clock starts. This allows the SLEEP instruction to be executed before starting a conversion.

The analog reference voltage is software selectable to either the device's positive and negative supply voltage (AVDD and AVSS), or the voltage level on the RA3/AN3/VREF+ and RA2/AN2/VREF- pins.

The A/D Converter has a unique feature of being able to operate while the device is in Sleep mode. To operate in Sleep, the A/D conversion clock must be derived from the A/D's Internal RC oscillator.

The output of the sample and hold is the input into the converter, which generates the result via successive approximation.

Each port pin associated with the A/D Converter can be configured as an analog input or as a digital I/O. The ADRESH and ADRESL registers contain the result of

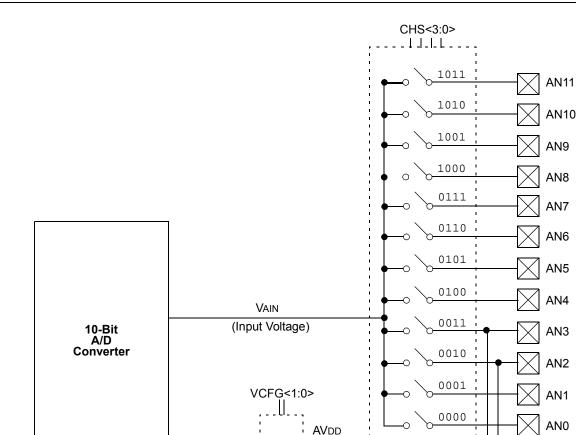
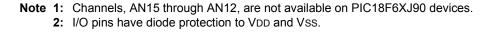


FIGURE 21-1: A/D BLOCK DIAGRAM^(1,2)

the A/D conversion. When the A/D conversion is complete, the result is loaded into the ADRESH:ADRESL register pair, the GO/DONE bit (ADCON0<1>) is cleared and the A/D Interrupt Flag bit, ADIF, is set.

A device Reset forces all registers to their Reset state. This forces the A/D module to be turned off and any conversion in progress is aborted. The value in the ADRESH:ADRESL register pair is not modified for a Power-on Reset. These registers will contain unknown data after a Power-on Reset.

The block diagram of the A/D module is shown in Figure 21-1.



O

 \cap

0

AVss

VREF+

VREF-

Reference

Voltage

L

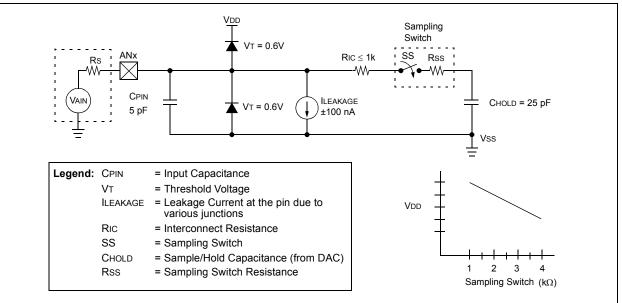
After the A/D module has been configured as desired, the selected channel must be acquired before the conversion is started. The analog input channels must have their corresponding TRIS bits selected as inputs. To determine acquisition time, see **Section 21.1 "A/D Acquisition Requirements"**. After this acquisition time has elapsed, the A/D conversion can be started. An acquisition time can be programmed to occur between setting the GO/DONE bit and the actual start of the conversion.

The following steps should be followed to do an A/D conversion:

- 1. Configure the A/D module:
 - Configure analog pins, voltage reference and digital I/O (ADCON1)
 - Select A/D input channel (ADCON0)
 - Select A/D acquisition time (ADCON2)
 - Select A/D conversion clock (ADCON2)
 - Turn on A/D module (ADCON0)
- 2. Configure A/D interrupt (if desired):
 - Clear the ADIF bit
 - · Set the ADIE bit
 - · Set the GIE bit



- 3. Wait the required acquisition time (if required).
- 4. Start conversion:
 Set GO/DONE bit (ADCON0<1>)
- 5. Wait for the A/D conversion to complete, by either:
 - Polling for the GO/DONE bit to be cleared OR
 - Waiting for the A/D interrupt
- 6. Read A/D Result registers (ADRESH:ADRESL); clear the ADIF bit, if required.
- 7. For the next conversion, go to step 1 or step 2, as required. The A/D conversion time per bit is defined as TAD. A minimum wait of 2 TAD is required before next acquisition starts.



21.1 A/D Acquisition Requirements

For the A/D Converter to meet its specified accuracy, the charge holding capacitor (CHOLD) must be allowed to fully charge to the input channel voltage level. The analog input model is shown in Figure 21-2. The source impedance (Rs) and the internal sampling switch (Rss) impedance directly affect the time required to charge the capacitor, CHOLD. The sampling switch (Rss) impedance varies over the device voltage (VDD). The source impedance affects the offset voltage at the analog input (due to pin leakage current). The maximum recommended impedance for analog sources is 2.5 k Ω . After the analog input channel is selected (changed), the channel must be sampled for at least the minimum acquisition time before starting a conversion.

Note: When the conversion is started, the holding capacitor is disconnected from the input pin.

EQUATION 21-1: A/D ACQUISITION TIME

TACQ = Amplifier Settling Time + Holding Capacitor Charging Time + Temperature Coefficient = TAMP + TC + TCOFF

EQUATION 21-2: A/D MINIMUM CHARGING TIME

VHOLD	=	$(\text{VREF} - (\text{VREF}/2048)) \bullet (1 - e^{(-\text{TC/CHOLD}(\text{RIC} + \text{Rss} + \text{Rs}))})$
or TC	=	-(CHOLD)(RIC + RSS + RS) ln(1/2048)

EQUATION 21-3: CALCULATING THE MINIMUM REQUIRED ACQUISITION TIME

TACQ	=	TAMP + TC + TCOFF
TAMP	=	0.2 μs
TCOFF	=	(Temp – 25°C)(0.02 μs/°C) (85°C – 25°C)(0.02 μs/°C) 1.2 μs
Tempera	ature c	oefficient is only required for temperatures $> 25^{\circ}$ C. Below 25° C, TCOFF = 0 ms.
Тс	=	-(CHOLD)(RIC + RSS + RS) $\ln(1/2048) \mu s$ -(25 pF) (1 k Ω + 2 k Ω + 2.5 k Ω) $\ln(0.0004883) \mu s$ 1.05 μs
TACQ	=	0.2 μs + 1 μs + 1.2 μs 2.4 μs

To calculate the minimum acquisition time, Equation 21-1 may be used. This equation assumes that 1/2 LSb error is used (1024 steps for the A/D). The 1/2 LSb error is the maximum error allowed for the A/D to meet its specified resolution.

Equation 21-3 shows the calculation of the minimum required acquisition time, TACQ. This calculation is based on the following application system assumptions:

CHOLD	=	25 pF
Rs	=	2.5 kΩ
Conversion Error	\leq	1/2 LSb
Vdd	=	$3V \rightarrow Rss = 2 k\Omega$
Temperature	=	85°C (system max.)

21.2 Selecting and Configuring Automatic Acquisition Time

The ADCON2 register allows the user to select an acquisition time that occurs each time the GO/DONE bit is set.

When the GO/DONE bit is set, sampling is stopped and a conversion begins. The user is responsible for ensuring the required acquisition time has passed between selecting the desired input channel and setting the GO/DONE bit. This occurs when the ACQT<2:0> bits (ADCON2<5:3>) remain in their Reset state ('000') and is compatible with devices that do not offer programmable acquisition times.

If desired, the ACQT bits can be set to select a programmable acquisition time for the A/D module. When the GO/DONE bit is set, the A/D module continues to sample the input for the selected acquisition time, then automatically begins a conversion. Since the acquisition time is programmed, there may be no need to wait for an acquisition time between selecting a channel and setting the GO/DONE bit.

In either case, when the conversion is completed, the GO/DONE bit is cleared, the ADIF flag is set and the A/D begins sampling the currently selected channel again. If an acquisition time is programmed, there is nothing to indicate if the acquisition time has ended or if the conversion has begun.

21.3 Selecting the A/D Conversion Clock

The A/D conversion time per bit is defined as TAD. The A/D conversion requires 11 TAD per 10-bit conversion. The source of the A/D conversion clock is software selectable.

There are seven possible options for TAD:

- 2 Tosc
- 4 Tosc
- 8 Tosc
- 16 Tosc
- 32 Tosc
- 64 Tosc
- Internal RC Oscillator

For correct A/D conversions, the A/D conversion clock (TAD) must be as short as possible, but greater than the minimum, TAD (see parameter 130 in Table 28-25 for more information).

Table 21-1 shows the resultant TAD times derived from the device operating frequencies and the A/D clock source selected.

TABLE 21-1: TAD vs. DEVICE OPERATING FREQUENCIES

AD Clock S	Maximum	
Operation	ADCS<2:0>	Device Frequency
2 Tosc	000	2.86 MHz
4 Tosc	100	5.71 MHz
8 Tosc	001	11.43 MHz
16 Tosc	101	22.86 MHz
32 Tosc	010	40.0 MHz
64 Tosc	110	40.0 MHz
RC ⁽²⁾	x11	1.00 MHz ⁽¹⁾

Note 1: The RC source has a typical TAD time of $4 \ \mu s$.

2: For device frequencies above 1 MHz, the device must be in Sleep mode for the entire conversion or the A/D accuracy may be out of specification.

21.4 Configuring Analog Port Pins

The ADCON1, TRISA, TRISF and TRISH registers control the operation of the A/D port pins. The port pins needed as analog inputs must have their corresponding TRIS bits set (input). If the TRIS bit is cleared (output), the digital output level (VOH or VOL) will be converted.

The A/D operation is independent of the state of the CHS<3:0> bits and the TRIS bits.

- Note 1: When reading the PORT register, all pins configured as analog input channels will read as cleared (a low level). Pins configured as digital inputs will convert an analog input. Analog levels on a digitally configured input will be accurately converted.
 - 2: Analog levels on any pin defined as a digital input may cause the digital input buffer to consume current out of the device's specification limits.

21.5 A/D Conversions

Figure 21-3 shows the operation of the A/D Converter after the GO/DONE bit has been set and the ACQT<2:0> bits are cleared. A conversion is started after the following instruction to allow entry into Sleep mode before the conversion begins.

Figure 21-4 shows the operation of the A/D Converter after the GO/DONE bit has been set, the ACQT<2:0> bits are set to '010' and selecting a 4 TAD acquisition time before the conversion starts.

Clearing the GO/DONE bit during a conversion will abort the current conversion. The A/D Result register pair will NOT be updated with the partially completed A/D conversion sample. This means the ADRESH:ADRESL registers will continue to contain the value of the last completed conversion (or the last value written to the ADRESH:ADRESL registers).

After the A/D conversion is completed or aborted, a 2 TAD wait is required before the next acquisition can be started. After this wait, acquisition on the selected channel is automatically started.

Note:	The GO/DONE bit should NOT be set in
	the same instruction that turns on the A/D.

21.6 Use of the CCP2 Trigger

An A/D conversion can be started by the "Special Event Trigger" of the CCP2 module. This requires that the CCP2M<3:0> bits (CCP2CON<3:0>) be programmed as '1011' and that the A/D module is enabled (ADON bit is set). When the trigger occurs, the GO/DONE bit will be set, starting the A/D acquisition and conversion, and the Timer1 (or Timer3) counter will be reset to zero. Timer1 (or Timer3) is reset to automatically repeat the A/D acquisition period with minimal software overhead (moving ADRESH:ADRESL to the desired location). The appropriate analog input channel must be selected and the minimum acquisition period is either timed by the user, or an appropriate TACQ time is selected before the Special Event Trigger sets the GO/DONE bit (starts a conversion).

If the A/D module is not enabled (ADON is cleared), the Special Event Trigger will be ignored by the A/D module but will still reset the Timer1 (or Timer3) counter.

FIGURE 21-3: A/D CONVERSION TAD CYCLES (ACQT<2:0> = 000, TACQ = 0)

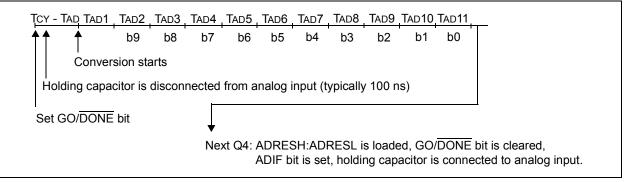
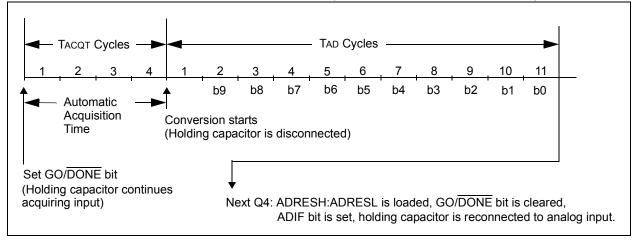


FIGURE 21-4: A/D CONVERSION TAD CYCLES (ACQT<2:0> = 010, TACQ = 4 TAD)



21.7 A/D Converter Calibration

The A/D Converter, in the PIC18F87J90 family of devices, includes a self-calibration feature which compensates for any offset generated within the module. The calibration process is automated and is initiated by setting the ADCAL bit (ADCON0<7>). The next time the GO/DONE bit is set, the module will perform a "dummy" conversion (that is, with reading none of the input channels) and store the resulting value internally to compensate for the offset. Thus, subsequent offsets will be compensated.

The calibration process assumes that the device is in a relatively steady-state operating condition. If A/D calibration is used, it should be performed after each device Reset or if there are other major changes in operating conditions.

21.8 Operation in Power-Managed Modes

The selection of the automatic acquisition time and A/D conversion clock is determined, in part, by the clock source and frequency while in a power-managed mode.

If the A/D is expected to operate while the device is in a power-managed mode, the ACQT<2:0> and ADCS<2:0> bits in ADCON2 should be updated in accordance with the power-managed mode clock that will be used. After the power-managed mode is entered (either of the power-managed Run modes), an A/D acquisition or conversion may be started. Once an acquisition or conversion is started, the device should continue to be clocked by the same power-managed mode clock source until the conversion has been completed. If desired, the device may be placed into the corresponding power-managed Idle mode during the conversion.

If the power-managed mode clock frequency is less than 1 MHz, the A/D RC clock source should be selected.

Operation in Sleep mode requires the A/D RC clock to be selected. If bits, ACQT<2:0>, are set to '000' and a conversion is started, the conversion will be delayed one instruction cycle to allow execution of the SLEEP instruction and entry to Sleep mode. The IDLEN and SCSx bits in the OSCCON register must have already been cleared prior to starting the conversion.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59
PIR1	—	ADIF	RC1IF	TX1IF	SSPIF	—	TMR2IF	TMR1IF	62
PIE1	_	ADIE	RC1IE	TX1IE	SSPIE	_	TMR2IE	TMR1IE	62
IPR1		ADIP	RC1IP	TX1IP	SSPIP	—	TMR2IP	TMR1IP	62
PIR3	—	LCDIF	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF	62
PIE3		LCDIE	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE	62
IPR3		LCDIP	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP	62
ADRESH	A/D Result	t Register Hi	gh Byte						61
ADRESL	A/D Result	t Register Lo	w Byte						61
ADCON0	ADCAL	—	CHS3	CHS2	CHS1	CHS0	GO/DONE	ADON	61
ADCON1	TRIGSEL		VCFG1	VCFG0	PCFG3	PCFG2	PCFG1	PCFG0	61
ADCON2	ADFM	_	ACQT2	ACQT1	ACQT0	ADCS2	ADCS1	ADCS0	61
CCP2CON		—	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	64
PORTA	RA7 ⁽¹⁾	RA6 ⁽¹⁾	RA5	RA4	RA3	RA2	RA1	RA0	63
TRISA	TRISA7 ⁽¹⁾	TRISA6 ⁽¹⁾	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	62
PORTF	RF7	RF6	RF5	RF4	RF3	RF2	RF1		62
TRISF	TRISF5	TRISF4	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1	_	62

TABLE 21-2: SUMMARY OF A/D REGISTERS

Legend: — = unimplemented, read as '0'. Shaded cells are not used for A/D conversion.

Note 1: RA<7:6> and their associated latch and direction bits are configured as port pins only when the internal oscillator is selected as the default clock source (FOSC2 Configuration bit = 0); otherwise, they are disabled and these bits read as '0'.

NOTES:

22.0 COMPARATOR MODULE

The analog comparator module contains two comparators that can be configured in a variety of ways. The inputs can be selected from the analog inputs multiplexed with pins, RF1 through RF6, as well as the on-chip voltage reference (see Section 23.0 "Comparator Voltage Reference Module"). The digital outputs (normal or inverted) are available at the pin level and can also be read through the control register. The CMCON register (Register 22-1) selects the comparator input and output configuration. Block diagrams of the various comparator configurations are shown in Figure 22-1.

REGISTER 22-1: CMCON: COMPARATOR MODULE CONTROL REGISTER

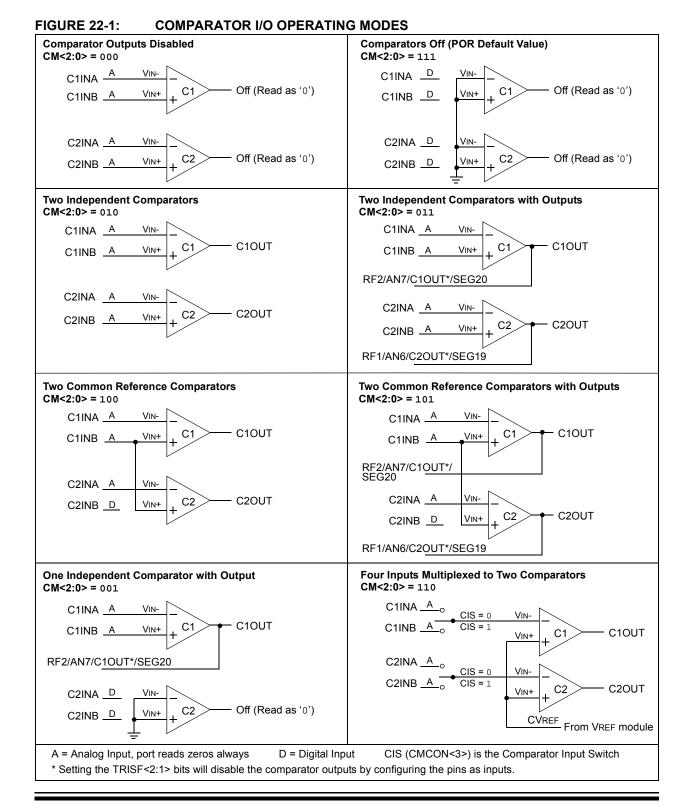
R-0	R-0	R/W-0	R/W-0	R/W-0	R/W-1	R/W-1	R/W-1	
C2OUT	C1OUT	C2INV	C1INV	CIS	CM2	CM1	CM0	
bit 7							bit 0	
Legend:								
R = Readable	R = Readable bit W = Writable bit		bit	U = Unimplemented bit, read as '0'				
-n = Value at POR		'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unknown		

bit 7	C2OUT : Comparator 2 Output bit $\frac{When C2INV = 0}{1 = C2 VIN+ > C2 VIN-}$ 0 = C2 VIN+ < C2 VIN- $\frac{When C2INV = 1}{1 = C2 VIN+ < C2 VIN-}$ 0 = C2 VIN+ < C2 VIN- 0 = C2 VIN+ > C2 VIN-
bit 6	C1OUT: Comparator 1 Output bit $\frac{When C1INV = 0}{1 = C1 VIN+ > C1 VIN-}$ 0 = C1 VIN+ < C1 VIN- $\frac{When C1INV = 1}{1 = C1 VIN+ < C1 VIN-}$ 0 = C1 VIN+ < C1 VIN-
bit 5	C2INV : Comparator 2 Output Inversion bit 1 = C2 output inverted 0 = C2 output not inverted
bit 4	C1INV: Comparator 1 Output Inversion bit 1 = C1 output inverted 0 = C1 output not inverted
bit 3	CIS: Comparator Input Switch bit <u>When CM<2:0> = 110:</u> 1 = C1 VIN- connects to RF5/AN10/CVREF/SEG23/C1INB C2 VIN- connects to RF3/AN8/SEG21/C2INB 0 = C1 VIN- connects to RF6/AN11/SEG24/C1INA C2 VIN- connects to RF4/AN9/SEG22/C2INA
bit 2-0	CM<2:0> : Comparator Mode bits Figure 22-1 shows the Comparator modes and the CM<2:0> bit settings.

22.1 Comparator Configuration

There are eight modes of operation for the comparators, shown in Figure 22-1. Bits, CM<2:0>, of the CMCON register are used to select these modes. The TRISF register controls the data direction of the comparator pins for each mode. If the Comparator mode is changed, the comparator output level may not be valid for the specified mode change delay shown in **Section 28.0 "Electrical Characteristics"**.

Note: Comparator interrupts should be disabled during a Comparator mode change; otherwise, a false interrupt may occur.



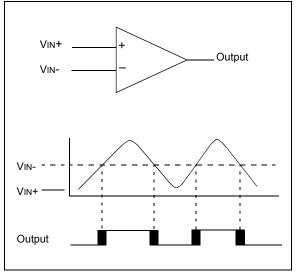
22.2 Comparator Operation

A single comparator is shown in Figure 22-2, along with the relationship between the analog input levels and the digital output. When the analog input at VIN+ is less than the analog input, VIN-, the output of the comparator is a digital low level. When the analog input at VIN+ is greater than the analog input, VIN-, the output of the comparator is a digital high level. The shaded areas of the output of the comparator in Figure 22-2 represent the uncertainty due to input offsets and response time.

22.3 Comparator Reference

Depending on the comparator operating mode, either an external or internal voltage reference may be used. The analog signal present at VIN- is compared to the signal at VIN+ and the digital output of the comparator is adjusted accordingly (Figure 22-2).





22.3.1 EXTERNAL REFERENCE SIGNAL

When external voltage references are used, the comparator module can be configured to have the comparators operate from the same or different reference sources. However, threshold detector applications may require the same reference. The reference signal must be between Vss and VDD and can be applied to either pin of the comparator(s).

22.3.2 INTERNAL REFERENCE SIGNAL

The comparator module also allows the selection of an internally generated voltage reference from the comparator voltage reference module. This module is described in more detail in **Section 23.0 "Comparator Voltage Reference Module"**.

The internal reference is only available in the mode where four inputs are multiplexed to two comparators (CM<2:0> = 110). In this mode, the internal voltage reference is applied to the VIN+ pin of both comparators.

22.4 Comparator Response Time

Response time is the minimum time, after selecting a new reference voltage or input source, before the comparator output has a valid level. If the internal reference is changed, the maximum delay of the internal voltage reference must be considered when using the comparator outputs. Otherwise, the maximum delay of the comparators should be used (see Section 28.0 "Electrical Characteristics").

22.5 Comparator Outputs

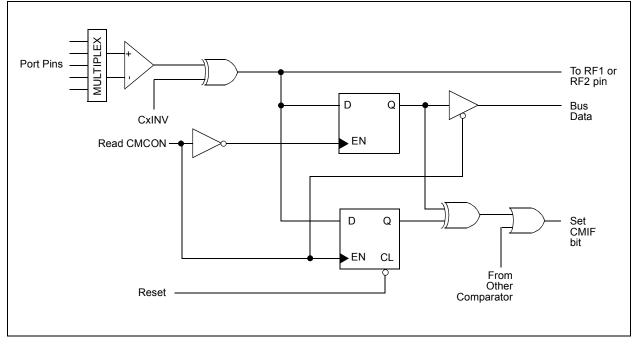
The comparator outputs are read through the CMCON register. These bits are read-only. The comparator outputs may also be directly output to the RF1 and RF2 I/O pins. When enabled, multiplexors in the output path of the RF1 and RF2 pins will switch and the output of each pin will be the unsynchronized output of the comparator. The uncertainty of each of the comparators is related to the input offset voltage and the response time given in the specifications. Figure 22-3 shows the comparator output block diagram.

The TRISF bits will still function as an output enable/ disable for the RF1 and RF2 pins while in this mode.

The polarity of the comparator outputs can be changed using the C2INV and C1INV bits (CMCON<5:4>).

- Note 1: When reading the PORT register, all pins configured as analog inputs will read as '0'. Pins configured as digital inputs will convert an analog input according to the Schmitt Trigger input specification.
 - Analog levels on any pin defined as a digital input may cause the input buffer to consume more current than is specified.

FIGURE 22-3: COMPARATOR OUTPUT BLOCK DIAGRAM



22.6 Comparator Interrupts

The comparator interrupt flag is set whenever there is a change in the output value of either comparator. Software will need to maintain information about the status of the output bits, as read from CMCON<7:6>, to determine the actual change that occurred. The CMIF bit (PIR2<6>) is the Comparator Interrupt Flag. The CMIF bit must be reset by clearing it. Since it is also possible to write a '1' to this register, a simulated interrupt may be initiated.

Both the CMIE bit (PIE2<6>) and the PEIE bit (INTCON<6>) must be set to enable the interrupt. In addition, the GIE bit (INTCON<7>) must also be set. If any of these bits are clear, the interrupt is not enabled, though the CMIF bit will still be set if an interrupt condition occurs.

Note:	If a change in the CMCON register					
	(C1OUT or C2OUT) should occur when a					
	read operation is being executed (start of					
	the Q2 cycle), then the CMIF (PIR2<6>)					
	interrupt flag may not get set.					

The user, in the Interrupt Service Routine, can clear the interrupt in the following manner:

- a) Any read or write of CMCON will end the mismatch condition.
- b) Clear flag bit, CMIF.

A mismatch condition will continue to set flag bit, CMIF. Reading CMCON will end the mismatch condition and allow flag bit, CMIF, to be cleared.

22.7 Comparator Operation During Sleep

When a comparator is active and the device is placed in Sleep mode, the comparator remains active and the interrupt is functional, if enabled. This interrupt will wake-up the device from Sleep mode, when enabled. Each operational comparator will consume additional current, as shown in the comparator specifications. To minimize power consumption while in Sleep mode, turn off the comparators (CM<2:0> = 111) before entering Sleep. If the device wakes up from Sleep, the contents of the CMCON register are not affected.

22.8 Effects of a Reset

A device Reset forces the CMCON register to its Reset state, causing the comparator modules to be turned off (CM<2:0> = 111). However, the input pins (RF3 through RF6) are configured as analog inputs by default on device Reset. The I/O configuration for these pins is determined by the setting of the PCFG<3:0> bits (ADCON1<3:0>). Therefore, device current is minimized when analog inputs are present at Reset time.

22.9 Analog Input Connection Considerations

A simplified circuit for an analog input is shown in Figure 22-4. Since the analog pins are connected to a digital output, they have reverse biased diodes to VDD and Vss. The analog input, therefore, must be between Vss and VDD. If the input voltage deviates from this

range by more than 0.6V in either direction, one of the diodes is forward biased and a latch-up condition may occur. A maximum source impedance of $10 \text{ k}\Omega$ is recommended for the analog sources. Any external component connected to an analog input pin, such as a capacitor or a Zener diode, should have very little leakage current.

FIGURE 22-4: COMPARATOR ANALOG INPUT MODEL

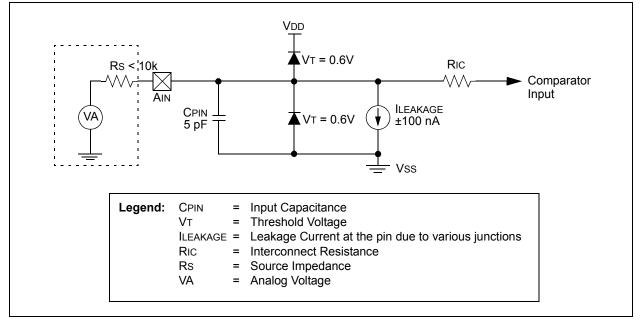


TABLE 22-1: REGISTERS ASSOCIATED WITH COMPARATOR MODULE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	59
PIR2	OSCFIF	CMIF		_	BCLIF	LVDIF	TMR3IF	_	62
PIE2	OSCFIE	CMIE		_	BCLIE	LVDIE	TMR3IE	_	62
IPR2	OSCFIP	CMIP	_		BCLIP	LVDIP	TMR3IP		62
CMCON	C2OUT	C10UT	C2INV	C1INV	CIS	CM2	CM1	CM0	61
CVRCON	CVREN	CVROE	CVRR	CVRSS	CVR3	CVR2	CVR1	CVR0	61
PORTF	RF7	RF6	RF5	RF4	RF3	RF2	RF1	_	62
LATF	LATF7	LATF6	LATF5	LATF4	LATF3	LATF2	LATF1		62
TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1	_	62

Legend: — = unimplemented, read as '0'. Shaded cells are unused by the comparator module.

NOTES:

23.0 COMPARATOR VOLTAGE REFERENCE MODULE

The comparator voltage reference is a 16-tap resistor ladder network that provides a selectable reference voltage. Although its primary purpose is to provide a reference for the analog comparators, it may also be used independently of them.

A block diagram of the module is shown in Figure 23-1. The resistor ladder is segmented to provide two ranges of CVREF values and has a power-down function to conserve power when the reference is not being used. The module's supply reference can be provided from either the device VDD/Vss or an external voltage reference.

23.1 Configuring the Comparator Voltage Reference

The comparator voltage reference module is controlled through the CVRCON register (Register 23-1). The comparator voltage reference provides two ranges of output voltage, each with 16 distinct levels. The range to be used is selected by the CVRR bit (CVRCON<5>). The primary difference between the ranges is the size of the steps selected by the CVREF Selection bits (CVR<3:0>), with one range offering finer resolution. The equations used to calculate the output of the comparator voltage reference are as follows:

<u>If CVRR = 1:</u> CVREF = ((CVR<3:0>)/24) x (CVRSRC) <u>If CVRR = 0:</u> CVREF = (CVRSRC/4) + ((CVR<3:0>)/32) x (CVRSRC)

The comparator reference supply voltage can come from either VDD and VSS, or the external VREF+ and VREF- that are multiplexed with RA2 and RA3. The voltage source is selected by the CVRSS bit (CVRCON<4>).

The settling time of the comparator voltage reference must be considered when changing the CVREF output (see Table 28-3 in **Section 28.0 "Electrical Characteristics"**).

REGISTER 23-1: CVRCON: COMPARATOR VOLTAGE REFERENCE CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
CVREN	CVROE ⁽¹⁾	CVRR	CVRSS	CVR3	CVR2	CVR1	CVR0
bit 7							bit 0

Legend:									
R = Reada	able bit	W = Writable bit	U = Unimplemented bit	, read as '0'					
-n = Value	at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown					
			—						
bit 7		Comparator Voltage Referen	ce Enable bit						
		EF circuit powered on							
		EF circuit powered down							
bit 6	CVROE:	Comparator VREF Output En	able bit ⁽¹⁾						
	1 = CVR	EF voltage level is also outpu	voltage level is also output on the RF5/AN10/CVREF/SEG23/C1INB pin						
	0 = CVR	EF voltage is disconnected from	oltage is disconnected from the RF5/AN10/CVREF/SEG23/C1INB pin						
bit 5	CVRR: C	CVRR: Comparator VREF Range Selection bit							
	1 = 0 to	= 0 to 0.667 CVRSRC, with CVRSRC/24 step size (low range)							
	0 = 0.25	CVRSRC to 0.75 CVRSRC, with	th CVRSRC/32 step size (high	range)					
bit 4	CVRSS:	Comparator VREF Source Se	lection bit						
	1 = Com	1 = Comparator reference source, CVRSRC = (VREF+) – (VREF-)							
		parator reference source, C\	. , . ,						
bit 3-0	CVR<3:0	Comparator VREF Value S	Selection bits ($0 \le (CVR < 3:0>)$	≤ 15)					
		When $CVRR = 1$:							
		((CVR<3:0>)/24) • (CVRSRC)							
	When C\								

Note 1: CVROE overrides the TRISF<5> bit setting.

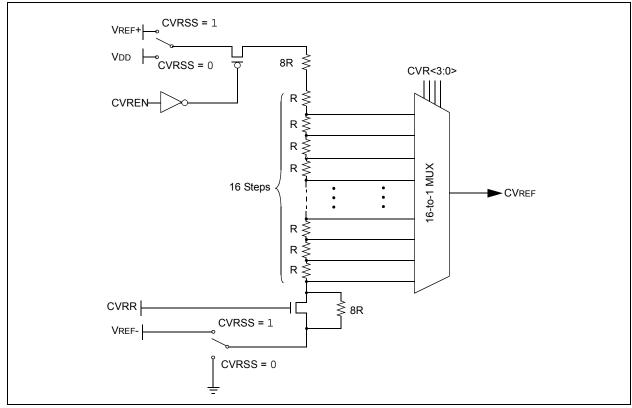


FIGURE 23-1: COMPARATOR VOLTAGE REFERENCE BLOCK DIAGRAM

23.2 Voltage Reference Accuracy/Error

The full range of voltage reference cannot be realized due to the construction of the module. The transistors on the top and bottom of the resistor ladder network (Figure 23-1) keep CVREF from approaching the reference source rails. The voltage reference is derived from the reference source; therefore, the CVREF output changes with fluctuations in that source. The tested absolute accuracy of the voltage reference can be found in **Section 28.0 "Electrical Characteristics"**.

23.3 Operation During Sleep

When the device wakes up from Sleep through an interrupt or a Watchdog Timer time-out, the contents of the CVRCON register are not affected. To minimize current consumption in Sleep mode, the voltage reference should be disabled.

23.4 Effects of a Reset

A device Reset disables the voltage reference by clearing bit, CVREN (CVRCON<7>). This Reset also disconnects the reference from the RA2 pin by clearing bit, CVROE (CVRCON<6>), and selects the high-voltage range by clearing bit, CVRR (CVRCON<5>). The CVR value select bits are also cleared.

23.5 Connection Considerations

The voltage reference module operates independently of the comparator module. The output of the reference generator may be connected to the RF5 pin if the CVROE bit is set. Enabling the voltage reference output onto RA2 when it is configured as a digital input will increase current consumption. Connecting RF5 as a digital output with CVRSS enabled will also increase current consumption.

The RF5 pin can be used as a simple D/A output with limited drive capability. Due to the limited current drive capability, a buffer must be used on the voltage reference output for external connections to VREF. Figure 23-2 shows an example buffering technique.

FIGURE 23-2: COMPARATOR VOLTAGE REFERENCE OUTPUT BUFFER EXAMPLE

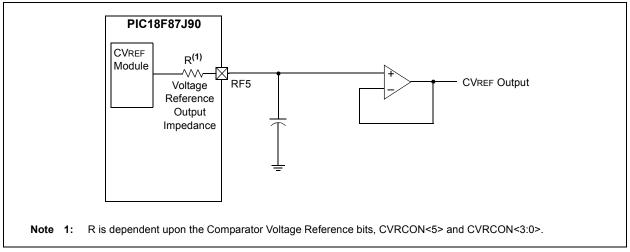


TABLE 23-1: REGISTERS ASSOCIATED WITH COMPARATOR VOLTAGE REFERENCE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
CVRCON	CVREN	CVROE	CVRR	CVRSS	CVR3	CVR2	CVR1	CVR0	61
CMCON	C2OUT	C10UT	C2INV	C1INV	CIS	CM2	CM1	CM0	61
TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1		62

Legend: — = unimplemented, read as '0'. Shaded cells are not used with the comparator voltage reference.

NOTES:

24.0 CHARGE TIME MEASUREMENT UNIT (CTMU)

The Charge Time Measurement Unit (CTMU) is a flexible analog module that provides accurate differential time measurement between pulse sources, as well as asynchronous pulse generation. By working with other on-chip analog modules, the CTMU can be used to precisely measure time, measure capacitance, measure relative changes in capacitance or generate output pulses with a specific time delay. The CTMU is ideal for interfacing with capacitive-based sensors.

The module includes the following key features:

- Up to 13 channels available for capacitive or time measurement input
- · On-chip precision current source
- Four-edge input trigger sources
- · Polarity control for each edge source

- Control of edge sequence
- · Control of response to edges
- · Time measurement resolution of 1 nanosecond
- High-precision time measurement
- Time delay of external or internal signal asynchronous to the system clock
- Accurate current source suitable for capacitive measurement

The CTMU works in conjunction with the A/D Converter to provide up to 13 channels for time or charge measurement, depending on the specific device and the number of A/D channels available. When configured for time delay, the CTMU is connected to one of the analog comparators. The level-sensitive input edge sources can be selected from four sources: two external inputs or the CCP1/CCP2 Special Event Triggers.

Figure 24-1 provides a block diagram of the CTMU.

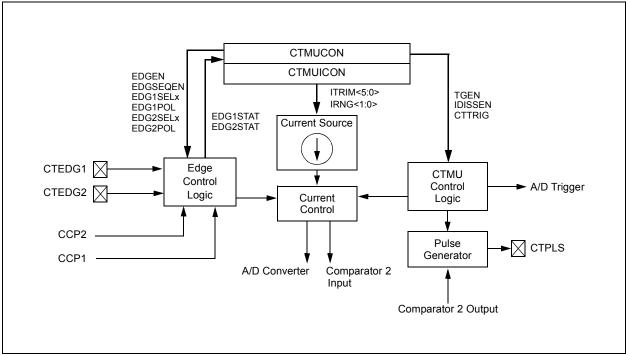


FIGURE 24-1: CTMU BLOCK DIAGRAM

24.1 CTMU Operation

The CTMU works by using a fixed current source to charge a circuit. The type of circuit depends on the type of measurement being made. In the case of charge measurement, the current is fixed and the amount of time the current is applied to the circuit is fixed. The amount of voltage read by the A/D is then a measurement of the capacitance of the circuit. In the case of time measurement, the current, as well as the capacitance of the circuit, are fixed. In this case, the voltage read by the A/D is then representative of the amount of time elapsed from the time the current source starts and stops charging the circuit.

If the CTMU is being used as a time delay, both capacitance and current source are fixed, as well as the voltage supplied to the comparator circuit. The delay of a signal is determined by the amount of time it takes the voltage to charge to the comparator threshold voltage.

24.1.1 THEORY OF OPERATION

The operation of the CTMU is based on the equation for charge:

$$C = I \bullet \frac{dV}{dT}$$

More simply, the amount of charge measured in coulombs in a circuit is defined as current in amperes (I) multiplied by the amount of time in seconds that the current flows (t). Charge is also defined as the capacitance in farads (C) multiplied by the voltage of the circuit (V). It follows that:

$$\mathbf{I} \bullet \mathbf{t} = \mathbf{C} \bullet \mathbf{V}$$

The CTMU module provides a constant, known current source. The A/D Converter is used to measure (V) in the equation, leaving two unknowns: capacitance (C) and time (t). The above equation can be used to calculate capacitance or time, by either relationship using the known fixed capacitance of the circuit:

$$\mathbf{t} = (\mathbf{C} \bullet \mathbf{V}) / \mathbf{I}$$

or by:

$$C = (I \bullet t)/V$$

using a fixed time that the current source is applied to the circuit.

24.1.2 CURRENT SOURCE

At the heart of the CTMU is a precision current source, designed to provide a constant reference for measurements. The level of current is user-selectable across three ranges or a total of two orders of magnitude, with the ability to trim the output in $\pm 2\%$ increments (nominal). The current range is selected by the IRNG<1:0> bits (CTMUICON<1:0>), with a value of '00' representing the lowest range.

Current trim is provided by the ITRIM<5:0> bits (CTMUICON<7:2>). These six bits allow trimming of the current source in steps of approximately 2% per step. Note that half of the range adjusts the current source positively and the other half reduces the current source. A value of '000000' is the neutral position (no change). A value of '100000' is the maximum negative adjustment (approximately -62%) and '011111' is the maximum positive adjustment (approximately +62%).

24.1.3 EDGE SELECTION AND CONTROL

CTMU measurements are controlled by edge events occurring on the module's two input channels. Each channel, referred to as Edge 1 and Edge 2, can be configured to receive input pulses from one of the edge input pins (CTEDG1 and CTEDG2) or the CCPx Special Event Triggers. The input channels are levelsensitive, responding to the instantaneous level on the channel rather than a transition between levels. The inputs are selected using the EDG1SEL and EDG2SEL bit pairs (CTMUCONL<3:2, 6:5>).

In addition to source, each channel can be configured for event polarity using the EDGE2POL and EDGE1POL bits (CTMUCONL<7,4>). The input channels can also be filtered for an edge event sequence (Edge 1 occurring before Edge 2) by setting the EDGSEQEN bit (CTMUCONH<2>).

24.1.4 EDGE STATUS

The CTMUCON register also contains two status bits, EDG2STAT and EDG1STAT (CTMUCONL<1:0>). Their primary function is to show if an edge response has occurred on the corresponding channel. The CTMU automatically sets a particular bit when an edge response is detected on its channel. The level-sensitive nature of the input channels also means that the status bits become set immediately if the channel's configuration is changed and is the same as the channel's current state.

The module uses the edge status bits to control the current source output to external analog modules (such as the A/D Converter). Current is only supplied to external modules when only one (but not both) of the status bits is set, and shuts current off when both bits are either set or cleared. This allows the CTMU to measure current only during the interval between edges. After both status bits are set, it is necessary to clear them before another measurement is taken. Both bits should be cleared simultaneously, if possible, to avoid re-enabling the CTMU current source.

In addition to being set by the CTMU hardware, the edge status bits can also be set by software. This is also the user's application to manually enable or disable the current source. Setting either one (but not both) of the bits enables the current source. Setting or clearing both bits at once disables the source.

24.1.5 INTERRUPTS

The CTMU sets its interrupt flag (PIR3<2>) whenever the current source is enabled, then disabled. An interrupt is generated only if the corresponding interrupt enable bit (PIE3<2>) is also set. If edge sequencing is not enabled (i.e., Edge 1 must occur before Edge 2), it is necessary to monitor the edge status bits, and determine which edge occurred last and caused the interrupt.

24.2 CTMU Module Initialization

The following sequence is a general guideline used to initialize the CTMU module:

- 1. Select the current source range using the IRNG bits (CTMUICON<1:0>).
- 2. Adjust the current source trim using the ITRIM bits (CTMUICON<7:2>).
- 3. Configure the edge input sources for Edge 1 and Edge 2 by setting the EDG1SEL and EDG2SEL bits (CTMUCONL<3:2 and 6:5>).
- Configure the input polarities for the edge inputs using the EDG1POL and EDG2POL bits (CTMUCONL<4,7>). The default configuration is for negative edge polarity (high-to-low transitions).
- Enable edge sequencing using the EDGSEQEN bit (CTMUCONH<2>). By default, edge sequencing is disabled.
- 6. Select the operating mode (Measurement or Time Delay) with the TGEN bit. The default mode is Time/Capacitance Measurement.
- Configure the module to automatically trigger an A/D conversion, when the second edge event has occurred, using the CTTRIG bit (CTMUCONH<0>). The conversion trigger is disabled by default.
- Discharge the connected circuit by setting the IDISSEN bit (CTMUCONH<1>); after waiting a sufficient time for the circuit to discharge, clear IDISSEN.
- 9. Disable the module by clearing the CTMUEN bit (CTMUCONH<7>).
- 10. Clear the Edge Status bits, EDG2STAT and EDG1STAT (CTMUCONL<1:0>).
- 11. Enable both edge inputs by setting the EDGEN bit (CTMUCONH<3>).
- 12. Enable the module by setting the CTMUEN bit.

Depending on the type of measurement or pulse generation being performed, one or more additional modules may also need to be initialized and configured with the CTMU module:

- Edge Source Generation: In addition to the external edge input pins, CCPx Special Event Triggers can be used as edge sources for the CTMU.
- Capacitance or Time Measurement: The CTMU module uses the A/D Converter to measure the voltage across a capacitor that is connected to one of the analog input channels.
- Pulse Generation: When generating system clock independent output pulses, the CTMU module uses Comparator 2 and the associated comparator voltage reference.

24.3 Calibrating the CTMU Module

The CTMU requires calibration for precise measurements of capacitance and time, as well as for accurate time delay. If the application only requires measurement of a relative change in capacitance or time, calibration is usually not necessary. An example of this type of application would include a capacitive touch switch, in which the touch circuit has a baseline capacitance and the added capacitance of the human body changes the overall capacitance of a circuit.

If actual capacitance or time measurement is required, two hardware calibrations must take place: the current source needs calibration to set it to a precise current, and the circuit being measured needs calibration to measure and/or nullify all other capacitance other than that to be measured.

24.3.1 CURRENT SOURCE CALIBRATION

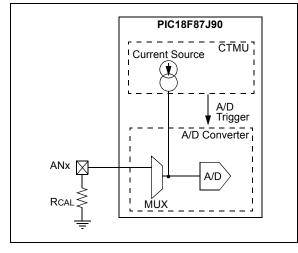
The current source on board the CTMU module has a range of $\pm 60\%$ nominal for each of three current ranges. Therefore, for precise measurements, it is possible to measure and adjust this current source by placing a high-precision resistor, RCAL, onto an unused analog channel. An example circuit is shown in Figure 24-2. The current source measurement is performed using the following steps:

- 1. Initialize the A/D Converter.
- 2. Initialize the CTMU.
- 3. Enable the current source by setting EDG1STAT (CTMUCONL<0>).
- 4. Issue settling time delay.
- 5. Perform A/D conversion.
- 6. Calculate the current source current using I = V/RCAL, where RCAL is a high-precision resistance and V is measured by performing an A/D conversion.

The CTMU current source may be trimmed with the trim bits in CTMUICON using an iterative process to get an exact desired current. Alternatively, the nominal value without adjustment may be used; it may be stored by the software for use in all subsequent capacitive or time measurements.

To calculate the value for *RCAL*, the nominal current must be chosen and then the resistance can be calculated. For example, if the A/D Converter reference voltage is 3.3V, use 70% of full scale or 2.31V as the desired approximate voltage to be read by the A/D Converter. If the range of the CTMU current source is selected to be 0.55 μ A, the resistor value needed is calculated as RCAL = 2.31V/0.55 μ A for a value of 4.2 M Ω . Similarly, if the current source is chosen to be 5.5 μ A, RCAL would be 420,000 Ω , and 42,000 Ω if the current source is set to 55 μ A.

FIGURE 24-2: CTMU CURRENT SOURCE CALIBRATION CIRCUIT



A value of 70% of full-scale voltage is chosen to make sure that the A/D Converter is in a range that is well above the noise floor. Keep in mind that if an exact current is chosen to incorporate the trimming bits from CTMUICON, the resistor value of RCAL may need to be adjusted accordingly. RCAL may be also adjusted to allow for available resistor values. RCAL should be of the highest precision available, keeping in mind the amount of precision needed for the circuit that the CTMU will be used to measure. A recommended minimum would be 0.1% tolerance.

The following examples show one typical method for performing a CTMU current calibration. Example 24-1 demonstrates how to initialize the A/D Converter and the CTMU; this routine is typical for applications using both modules. Example 24-2 demonstrates one method for the actual calibration routine. Note that this method manually triggers the A/D Converter. This is done to demonstrate the entire stepwise process. It is also possible to automatically trigger the conversion by setting the CTMU's CTTRIG bit (CTMUCONH<0>).

EXAMPLE 24-1: SETUP FOR CTMU CALIBRATION ROUTINES

```
#include "pl8cxxx.h"
void setup(void)
{ //CTMUCON - CTMU Control register
  CTMUCONH = 0 \times 00;
                         //make sure CTMU is disabled
  CTMUCONL = 0X90;
  //CTMU continues to run when emulator is stopped,CTMU continues
  //to run in idle mode, Time Generation mode disabled, Edges are blocked
  //No edge sequence order, Analog current source not grounded, trigger
  //output disabled, Edge2 polarity = positive level, Edge2 source =
  //source 0, Edgel polarity = positive level, Edgel source = source 0,
  // Set Edge status bits to zero
   //CTMUICON - CTMU Current Control Register
  CTMUICON = 0 \times 01;
                        //0.55uA, Nominal - No Adjustment
//Setup AD converter;
TRISA=0x04;
                         //set channel 2 as an input
  // ADCON2
  ADCON2bits.ADFM=1;
                        // Resulst format 1= Right justified
  ADCON2bits.ACQT=1;
                         // Acquition time 7 = 20TAD 2 = 4TAD 1=2TAD
                         // Clock conversion bits 6= FOSC/64 2=FOSC/32
  ADCON2bits.ADCS=2;
  // ADCON1
                        // Normal A/D conversion operation
  ADCON1bits.ADCAL=0;
  ADCON1bits.PCFG=0xC;
                         // Configures ANO to AN2 as analog
  // ADCONO
                         // Vref+ = AVdd
  ADCON0bits.VCFG0 =0;
  ADCON0bits.VCFG1 =0;
                         // Vref- = AVss
                         // Select ADC channel
  ADCON0bits.CHS=2;
  ADCON0bits.ADON=1;
                        // Turn on ADC
}
```

EXAMPLE 24-2: CURRENT CALIBRATION ROUTINE

```
#include "pl8cxxx.h"
#define COUNT 500
                                         //@ 8MHz = 125uS.
#define DELAY for(i=0;i<COUNT;i++)</pre>
#define RCAL .027
                                         //R value is 4200000 (4.2M)
                                         //scaled so that result is in
                                         //1/100th of uA
#define ADSCALE 1023
                                         //for unsigned conversion 10 sig bits
#define ADREF 3.3
                                         //Vdd connected to A/D Vr+
int main(void)
{
   int i;
   int j = 0; //index for loop
   unsigned int Vread = 0;
   double VTot = 0;
   float Vavg=0, Vcal=0, CTMUISrc = 0; //float values stored for calcs
//assume CTMU and A/D have been setup correctly
//see Example 25-1 for CTMU & A/D setup
setup();
CTMUCONHbits.CTMUEN = 1;
                                         //Enable the CTMU
   for(j=0;j<10;j++)</pre>
    {
                                         //drain charge on the circuit
       CTMUCONHbits.IDISSEN = 1;
       DELAY;
                                         //wait 125us
       CTMUCONHbits.IDISSEN = 0;
                                         //end drain of circuit
       CTMUCONLbits.EDG1STAT = 1;
                                         //Begin charging the circuit
                                         //using CTMU current source
                                         //wait for 125us
       DELAY;
                                         //Stop charging circuit
       CTMUCONLbits.EDG1STAT = 0;
       PIR1bits.ADIF = 0;
                                         //make sure A/D Int not set
       ADCON0bits.GO=1;
                                         //and begin A/D conv.
       while(!PIR1bits.ADIF);
                                         //Wait for A/D convert complete
       Vread = ADRES;
                                         //Get the value from the A/D
       PIR1bits.ADIF = 0;
                                         //Clear A/D Interrupt Flag
       VTot += Vread;
                                         //Add the reading to the total
   }
   Vavg = (float)(VTot/10.000);
                                         //Average of 10 readings
   Vcal = (float)(Vavg/ADSCALE*ADREF);
   CTMUISrc = Vcal/RCAL;
                                         //CTMUISrc is in 1/100ths of uA
```

24.3.2 CAPACITANCE CALIBRATION

There is a small amount of capacitance from the internal A/D Converter sample capacitor as well as stray capacitance from the circuit board traces and pads that affect the precision of capacitance measurements. A measurement of the stray capacitance can be taken by making sure the desired capacitance to be measured has been removed. The measurement is then performed using the following steps:

- 1. Initialize the A/D Converter and the CTMU.
- 2. Set EDG1STAT (= 1).
- 3. Wait for a fixed delay of time, t.
- 4. Clear EDG1STAT.
- 5. Perform an A/D conversion.
- 6. Calculate the stray and A/D sample capacitances:

$$COFFSET = CSTRAY + CAD = (I \cdot t)/V$$

where I is known from the current source measurement step, t is a fixed delay and V is measured by performing an A/D conversion.

This measured value is then stored and used for calculations of time measurement or subtracted for capacitance measurement. For calibration, it is expected that the capacitance of CSTRAY + CAD is approximately known; CAD is approximately 4 pF.

An iterative process may need to be used to adjust the time, t, that the circuit is charged to obtain a reasonable voltage reading from the A/D Converter. The value of t may be determined by setting COFFSET to a theoretical value, then solving for t. For example, if CSTRAY is theoretically calculated to be 11 pF, and V is expected to be 70% of VDD, or 2.31V, then t would be:

or 63 µs.

See Example 24-3 for a typical routine for CTMU capacitance calibration.

EXAMPLE 24-3: CAPACITANCE CALIBRATION ROUTINE

```
#include "pl8cxxx.h"
#define COUNT 25
                                          //@ 8MHz INTFRC = 62.5 us.
#define ETIME COUNT*2.5
                                          //time in uS
#define DELAY for(i=0;i<COUNT;i++)</pre>
#define ADSCALE 1023
                                          //for unsigned conversion 10 sig bits
#define ADREF 3.3
                                          //Vdd connected to A/D Vr+
#define RCAL .027
                                          //R value is 4200000 (4.2M)
                                          //scaled so that result is in
                                          //1/100th of uA
int main(void)
{
    int i;
   int j = 0;
                                          //index for loop
    unsigned int Vread = 0;
    float CTMUISrc, CTMUCap, Vavg, VTot, Vcal;
//\ensuremath{\mathsf{assume}} CTMU and A/D have been setup correctly
//see Example 25-1 for CTMU & A/D setup
setup();
CTMUCONHbits.CTMUEN = 1;
                                          //Enable the CTMU
    for(j=0;j<10;j++)</pre>
    {
        CTMUCONHbits.IDISSEN = 1;
                                          //drain charge on the circuit
       DELAY;
                                          //wait 125us
        CTMUCONHbits.IDISSEN = 0;
                                          //end drain of circuit
        CTMUCONLbits.EDG1STAT = 1;
                                          //Begin charging the circuit
                                          //using CTMU current source
        DELAY;
                                          //wait for 125us
        CTMUCONLbits.EDG1STAT = 0;
                                          //Stop charging circuit
        PIR1bits.ADIF = 0;
                                          //make sure A/D Int not set
        ADCON0bits.GO=1;
                                         //and begin A/D conv.
        while(!PIR1bits.ADIF);
                                         //Wait for A/D convert complete
        Vread = ADRES;
                                          //Get the value from the A/D
        PIR1bits.ADIF = 0;
                                          //Clear A/D Interrupt Flag
        VTot += Vread;
                                          //Add the reading to the total
    }
    Vavg = (float)(VTot/10.000);
                                          //Average of 10 readings
    Vcal = (float)(Vavg/ADSCALE*ADREF);
                                          //CTMUISrc is in 1/100ths of uA
    CTMUISrc = Vcal/RCAL;
    CTMUCap = (CTMUISrc*ETIME/Vcal)/100;
}
```

24.4 Measuring Capacitance with the CTMU

There are two separate methods of measuring capacitance with the CTMU. The first is the absolute method, in which the actual capacitance value is desired. The second is the relative method, in which the actual capacitance is not needed, rather an indication of a change in capacitance is required.

24.4.1 ABSOLUTE CAPACITANCE MEASUREMENT

For absolute capacitance measurements, both the current and capacitance calibration steps found in **Section 24.3 "Calibrating the CTMU Module"** should be followed. Capacitance measurements are then performed using the following steps:

- 1. Initialize the A/D Converter.
- 2. Initialize the CTMU.
- 3. Set EDG1STAT.
- 4. Wait for a fixed delay, T.
- 5. Clear EDG1STAT.
- 6. Perform an A/D conversion.
- Calculate the total capacitance, CTOTAL = (I * T)/V, where I is known from the current source measurement step (Section 24.3.1 "Current Source Calibration"), T is a fixed delay and V is measured by performing an A/D conversion.
- 8. Subtract the stray and A/D capacitance (COFFSET from Section 24.3.2 "Capacitance Calibration") from CTOTAL to determine the measured capacitance.

24.4.2 RELATIVE CHARGE MEASUREMENT

An application may not require precise capacitance measurements. For example, when detecting a valid press of a capacitance-based switch, detecting a relative change of capacitance is of interest. In this type of application, when the switch is open (or not touched), the total capacitance is the capacitance of the combination of the board traces, the A/D Converter, etc. A larger voltage will be measured by the A/D Converter. When the switch is closed (or is touched), the total capacitance is larger due to the addition of the capacitance of the human body to the above listed capacitances and a smaller voltage will be measured by the A/D Converter.

Detecting capacitance changes is easily accomplished with the CTMU using these steps:

- 1. Initialize the A/D Converter and the CTMU.
- 2. Set EDG1STAT.
- 3. Wait for a fixed delay.
- 4. Clear EDG1STAT.
- 5. Perform an A/D conversion.

The voltage measured by performing the A/D conversion is an indication of the relative capacitance. Note that in this case, no calibration of the current source or circuit capacitance measurement is needed. See Example 24-4 for a sample software routine for a capacitive touch switch.

EXAMPLE 24-4: ROUTINE FOR CAPACITIVE TOUCH SWITCH

```
#include "p18cxxx.h"
#define COUNT 500
                                         //@ 8MHz = 125uS.
#define DELAY for(i=0;i<COUNT;i++)</pre>
#define OPENSW 1000
                                         //Un-pressed switch value
#define TRIP 300
                                         //Difference between pressed
                                         //and un-pressed switch
#define HYST 65
                                         //amount to change
                                         //from pressed to un-pressed
#define PRESSED 1
#define UNPRESSED 0
int main(void)
{
   unsigned int Vread;
                                         //storage for reading
   unsigned int switchState;
   int i;
   //assume CTMU and A/D have been setup correctly
   //see Example 25-1 for CTMU & A/D setup
   setup();
   CTMUCONHbits.CTMUEN = 1;
                                        //Enable the CTMU
   CTMUCONHbits.IDISSEN = 1;
                                        //drain charge on the circuit
                                         //wait 125us
   DELAY;
   CTMUCONHbits.IDISSEN = 0;
                                         //end drain of circuit
   CTMUCONLbits.EDG1STAT = 1;
                                        //Begin charging the circuit
                                         //using CTMU current source
                                         //wait for 125us
   DELAY;
   CTMUCONLbits.EDG1STAT = 0;
                                         //Stop charging circuit
   PIR1bits.ADIF = 0;
                                         //make sure A/D Int not set
   ADCON0bits.GO=1;
                                         //and begin A/D conv.
   while(!PIR1bits.ADIF);
                                        //Wait for A/D convert complete
   Vread = ADRES;
                                        //Get the value from the A/D
   if(Vread < OPENSW - TRIP)
   {
        switchState = PRESSED;
    }
   else if(Vread > OPENSW - TRIP + HYST)
   {
       switchState = UNPRESSED;
   }
```

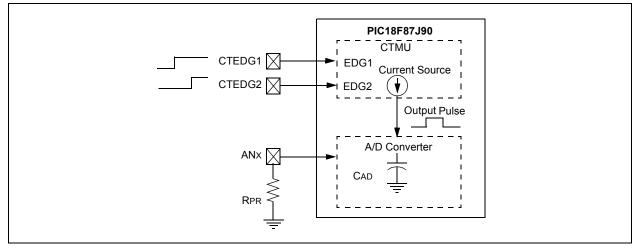
24.5 Measuring Time with the CTMU Module

Time can be precisely measured after the ratio (C/I) is measured from the current and capacitance calibration step by following these steps:

- 1. Initialize the A/D Converter and the CTMU.
- 2. Set EDG1STAT.
- 3. Set EDG2STAT.
- 4. Perform an A/D conversion.
- 5. Calculate the time between edges as T = (C/I) * V, where I is calculated in the current calibration step (Section 24.3.1 "Current Source Calibration"), C is calculated in the capacitance calibration step (Section 24.3.2 "Capacitance Calibration") and V is measured by performing the A/D conversion.

It is assumed that the time measured is small enough that the capacitance, COFFSET, provides a valid voltage to the A/D Converter. For the smallest time measurement, always set the A/D Channel Select register (AD1CHS) to an unused A/D channel; the corresponding pin for which is not connected to any circuit board trace. This minimizes added stray capacitance, keeping the total circuit capacitance close to that of the A/D Converter itself (25 pF). To measure longer time intervals, an external capacitor may be connected to an A/D channel and this channel selected when making a time measurement.

FIGURE 24-3: TYPICAL CONNECTIONS AND INTERNAL CONFIGURATION FOR TIME MEASUREMENT



24.6 Creating a Delay with the CTMU Module

A unique feature on board the CTMU module is its ability to generate system clock independent output pulses based on an external capacitor value. This is accomplished using the internal comparator voltage reference module, Comparator 2 input pin and an external capacitor. The pulse is output onto the CTPLS pin. To enable this mode, set the TGEN bit.

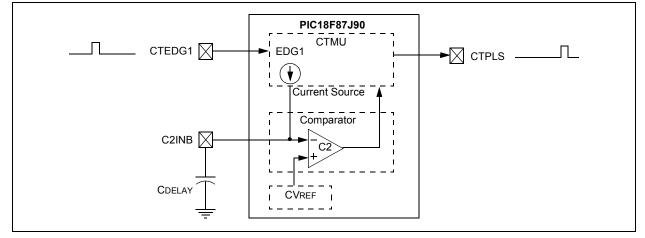
See Figure 24-4 for an example circuit. CPULSE is chosen by the user to determine the output pulse width on CTPLS. The pulse width is calculated by T = (CPULSE/I)*V, where I is known from the current source measurement step (Section 24.3.1 "Current Source Calibration") and V is the internal reference voltage (CVREF).

An example use of this feature is for interfacing with variable capacitive-based sensors, such as a humidity sensor. As the humidity varies, the pulse-width output on CTPLS will vary. The CTPLS output pin can be connected to an input capture pin and the varying pulse width is measured to determine the humidity in the application.

Follow these steps to use this feature:

- 1. Initialize Comparator 2.
- 2. Initialize the comparator voltage reference.
- 3. Initialize the CTMU and enable time delay generation by setting the TGEN bit.
- 4. Set EDG1STAT.
- 5. When CPULSE charges to the value of the voltage reference trip point, an output pulse is generated on CTPLS.

FIGURE 24-4: TYPICAL CONNECTIONS AND INTERNAL CONFIGURATION FOR PULSE DELAY GENERATION



24.7 Operation During Sleep/Idle Modes

24.7.1 SLEEP MODE AND DEEP SLEEP MODES

When the device enters any Sleep mode, the CTMU module current source is always disabled. If the CTMU is performing an operation that depends on the current source when Sleep mode is invoked, the operation may not terminate correctly. Capacitance and time measurements may return erroneous values.

24.7.2 IDLE MODE

The behavior of the CTMU in Idle mode is determined by the CTMUSIDL bit (CTMUCONH<5>). If CTMUSIDL is cleared, the module will continue to operate in Idle mode. If CTMUSIDL is set, the module's current source is disabled when the device enters Idle mode. If the module is performing an operation when Idle mode is invoked, in this case, the results will be similar to those with Sleep mode.

24.8 Effects of a Reset on CTMU

Upon Reset, all registers of the CTMU are cleared. This leaves the CTMU module disabled, its current source is turned off and all configuration options return to their default settings. The module needs to be re-initialized following any Reset.

If the CTMU is in the process of taking a measurement at the time of Reset, the measurement will be lost. A partial charge may exist on the circuit that was being measured and should be properly discharged before the CTMU makes subsequent attempts to make a measurement. The circuit is discharged by setting and then clearing the IDISSEN bit (CTMUCONH<1>) while the A/D Converter is connected to the appropriate channel.

24.9 Registers

There are three control registers for the CTMU:

- CTMUCONH
- CTMUCONL
- CTMUICON

The CTMUCONH and CTMUCONL registers (Register 24-1 and Register 24-2) contain control bits for configuring the CTMU module edge source selection, edge source polarity selection, edge sequencing, A/D trigger, analog circuit capacitor discharge and enables. The CTMUICON register (Register 24-3) has bits for selecting the current source range and current source trim.

REGISTER 24-1: CTMUCONH: CTMU CONTROL HIGH REGISTER

R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
CTMUEN	_	CTMUSIDL	TGEN	EDGEN	EDGSEQEN	IDISSEN	CTTRIG
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	CTMUEN: CTMU Enable bit
	1 = Module is enabled
	0 = Module is disabled
bit 6	Unimplemented: Read as '0'
bit 5	CTMUSIDL: Stop in Idle Mode bit
	 1 = Discontinue module operation when device enters Idle mode 0 = Continue module operation in Idle mode
bit 4	TGEN: Time Generation Enable bit
	1 = Enables edge delay generation
	0 = Disables edge delay generation
bit 3	EDGEN: Edge Enable bit
	1 = Edges are not blocked
	0 = Edges are blocked
bit 2	EDGSEQEN: Edge Sequence Enable bit
	1 = Edge 1 event must occur before Edge 2 event can occur
	0 = No edge sequence is needed
bit 1	IDISSEN: Analog Current Source Control bit
	1 = Analog current source output is grounded
	0 = Analog current source output is not grounded
bit 0	CTTRIG: Trigger Control bit
	1 = Trigger output is enabled
	0 = Trigger output is disabled

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
EDG2POL	EDG2SEL1	EDG2SEL0	EDG1POL	EDG1SEL1	EDG1SEL0	EDG2STAT	EDG1STAT
DG2POL	EDG25ELT	EDG25ELU	EDGIPOL	EDGISELI	EDGISELU	EDG25TAT	
							bit 0
Legend:							
R = Readable	bit	W = Writable	hit	II = I Inimplem	nented bit, read	l as 'O'	
-n = Value at		'1' = Bit is set		·0' = Bit is clea		x = Bit is unkr	
					area		10111
bit 7	EDG2POI : F	dge 2 Polarity	Select bit				
		rogrammed for		le response			
		rogrammed for					
bit 6-5	EDG2SEL<1:	: 0>: Edge 2 So	urce Select bit	s			
	11 = CTEDG	1 pin					
	10 = CTEDG						
		pecial Event Tr pecial Event Tr					
bit 4		dge 1 Polarity					
bit 4		rogrammed for		le response			
	• .	rogrammed for					
bit 3-2	EDG1SEL<1:	:0>: Edge 1 So	urce Select bit	S			
	11 = CTEDG	1 pin					
	10 = CTEDG						
		pecial Event Tr pecial Event Tr					
bit 1		Edge 2 Status b					
bit i		vent has occur					
	U U	vent has not or					
bit 0	EDG1STAT: E	Edge 1 Status b	oit				
		vent has occur					
	0 = Edge 1 e	vent has not or	curred				

REGISTER 24-2: CTMUCONL: CTMU CONTROL LOW REGISTER

REGISTER 24-3: CTMUICON: CTMU CURRENT CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
ITRIM5	ITRIM4	ITRIM3	ITRIM2	ITRIM1	ITRIM0	IRNG1	IRNG0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-2	ITRIM<5:0>: Current Source Trim bits 011111 = Maximum positive change from nominal current 011110
	000001 = Minimum positive change from nominal current 000000 = Nominal current output specified by IRNG<1:0> 111111 = Minimum negative change from nominal current
	100010 100001 = Maximum negative change from nominal current
bit 1-0	IRNG<1:0>: Current Source Range Select bits
	 11 = 100 x Base current 10 = 10 x Base current 01 = Base current level (0.55 μA nominal)

00 = Current source disabled

TABLE 24-1: REGISTERS ASSOCIATED WITH CTMU MODULE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page:
CTMUCONH	CTMUEN	_	CTMUSIDL	TGEN	EDGEN	EDGSEQEN	IDISSEN	CTTRIG	_
CTMUCONL	EDG2POL	EDG2SEL1	EDG2SEL0	EDG1POL	EDG1SEL1	EDG1SEL0	EDG2STAT	EDG1STAT	—
CTMUICON	ITRIM5	ITRIM4	ITRIM3	ITRIM2	ITRIM1	ITRIM0	IRNG1	IRNG0	_

Legend: — = unimplemented, read as '0'. Shaded cells are not used during ECCP operation.

NOTES:

25.0 SPECIAL FEATURES OF THE CPU

PIC18F87J90 family devices include several features intended to maximize reliability and minimize cost through elimination of external components. These are:

- Oscillator Selection
- Resets:
 - Power-on Reset (POR)
 - Power-up Timer (PWRT)
 - Oscillator Start-up Timer (OST)
 - Brown-out Reset (BOR)
- Interrupts
- Watchdog Timer (WDT)
- · Fail-Safe Clock Monitor
- Two-Speed Start-up
- Code Protection
- In-Circuit Serial Programming

The oscillator can be configured for the application depending on frequency, power, accuracy and cost. All of the options are discussed in detail in **Section 3.0 "Oscillator Configurations"**.

A complete discussion of device Resets and interrupts is available in previous sections of this data sheet.

In addition to their Power-up and Oscillator Start-up Timers provided for Resets, the PIC18F87J90 family of devices have a configurable Watchdog Timer which is controlled in software.

The inclusion of an Internal RC oscillator also provides the additional benefits of a Fail-Safe Clock Monitor (FSCM) and Two-Speed Start-up. FSCM provides for background monitoring of the peripheral clock and automatic switchover in the event of its failure. Two-Speed Start-up enables code to be executed almost immediately on start-up, while the primary clock source completes its start-up delays.

All of these features are enabled and configured by setting the appropriate Configuration register bits.

25.1 Configuration Bits

The Configuration bits can be programmed (read as '0') or left unprogrammed (read as '1') to select various device configurations. These bits are mapped, starting at program memory location, 300000h. A complete list is shown in Table 25-2. A detailed explanation of the various bit functions is provided in Register 25-1 through Register 25-6.

25.1.1 CONSIDERATIONS FOR CONFIGURING PIC18F87J90 FAMILY DEVICES

Devices of the PIC18F87J90 family do not use persistent memory registers to store configuration information. The configuration bytes are implemented as volatile memory which means that configuration data must be programmed each time the device is powered up.

Configuration data is stored in the four words at the top of the on-chip program memory space, known as the Flash Configuration Words. It is stored in program memory in the same order shown in Table 25-2, with CONFIG1L at the lowest address and CONFIG3H at the highest. The data is automatically loaded in the proper Configuration registers during device power-up.

When creating applications for these devices, users should always specifically allocate the location of the Flash Configuration Word for configuration data. This is to make certain that program code is not stored in this address when the code is compiled.

The volatile memory cells used for the Configuration bits always reset to '1' on Power-on Resets. For all other types of Reset events, the previously programmed values are maintained and used without reloading from program memory.

The four Most Significant bits of CONFIG1H, CONFIG2H and CONFIG3H in program memory should also be '1111'. This makes these Configuration Words appear to be NOP instructions in the remote event that their locations are ever executed by accident. Since Configuration bits are not implemented in the corresponding locations, writing '1's to these locations has no effect on device operation.

To prevent inadvertent configuration changes during code execution, all programmable Configuration bits are write-once. After a bit is initially programmed during a power cycle, it cannot be written to again. Changing a device configuration requires that power to the device be cycled.

	REGISTERS								
Configuration Byte	Code Space Address	Configuration Register Address							
CONFIG1L	XXXF8h	300000h							
CONFIG1H	XXXF9h	300001h							
CONFIG2L	XXXFAh	300002h							
CONFIG2H	XXXFBh	300003h							
CONFIG3L	XXXFCh	300004h							
CONFIG3H	XXXFDh	300005h							

TABLE 25-1: MAPPING OF THE FLASH CONFIGURATION WORDS TO THE CONFIGURATION REGISTERS

TABLE 25-2: CONFIGURATION BITS AND DEVICE IDs

File	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Default/ Unprogrammed Value ⁽¹⁾
300000h	CONFIG1L	DEBUG	XINST	STVREN	_	_	_	_	WDTEN	1111
300001h	CONFIG1H	(2)	(2)	(2)	(2)	_(3)	CP0	_	_	01
300002h	CONFIG2L	IESO	FCMEN	_	LPT1OSC	T1DIG	FOSC2	FOSC1	FOSC0	11-1 1111
300003h	CONFIG2H	(2)	(2)	(2)	(2)	WDTPS3	WDTPS2	WDTPS1	WDTPS0	1111
300004h	CONFIG3L	_(2)	(2)	(2)	_(2)	_	_	RTCOSC	_	1-
300005h	CONFIG3H	_(2)	(2)	(2)	(2)		_	_	CCP2MX	1
3FFFFEh	DEVID1	DEV2	DEV1	DEV0	REV4	REV3	REV2	REV1	REV0	xxxx xxxx ⁽⁴⁾
3FFFFFh	DEVID2	DEV10	DEV9	DEV8	DEV7	DEV6	DEV5	DEV4	DEV3	0000 10x1 ⁽⁴⁾

Legend: x = unknown, - = unimplemented. Shaded cells are unimplemented, read as '0'.

Note 1: Values reflect the unprogrammed state as received from the factory and following Power-on Resets. In all other Reset states, the configuration bytes maintain their previously programmed states.

2: The value of these bits in program memory should always be 'l'. This ensures that the location is executed as a NOP if it is accidentally executed.

3: This bit should always be maintained as '0'.

4: See Register 25-7 and Register 25-8 for DEVID values. These registers are read-only and cannot be programmed by the user.

REGISTER 25-1: CONFIG1L: CONFIGURATION REGISTER 1 LOW (BYTE ADDRESS 300000h)

R/WO-1	R/WO-1	R/WO-1	U-0	U-0	U-0	U-0	R/WO-1
DEBUG	XINST	STVREN	_	_	—	_	WDTEN
bit 7							bit 0

Legend:			
R = Readable bit	WO = Write-Once bit	U = Unimplemented b	pit, read as '0'
-n = Value when device i	s unprogrammed	'1' = Bit is set	'0' = Bit is cleared

bit 7	DEBUG: Background Debugger Enable bit
	1 = Background debugger disabled; RB6 and RB7 configured as general purpose I/O pins
	0 = Background debugger enabled; RB6 and RB7 are dedicated to In-Circuit Debug
bit 6	XINST: Extended Instruction Set Enable bit
	 1 = Instruction set extension and Indexed Addressing mode enabled 0 = Instruction set extension and Indexed Addressing mode disabled (Legacy mode)
bit 5	STVREN: Stack Overflow/Underflow Reset Enable bit
	 1 = Reset on stack overflow/underflow enabled 0 = Reset on stack overflow/underflow disabled
bit 4-1	Unimplemented: Read as '0'
bit 0	WDTEN: Watchdog Timer Enable bit
	1 = WDT enabled
	0 = WDT disabled (control is placed on SWDTEN bit)

REGISTER 25-2: CONFIG1H: CONFIGURATION REGISTER 1 HIGH (BYTE ADDRESS 300001h)

U-0	U-0	U-0	U-0	U-0	R/WO-1	U-0	U-0
(1)	(1)	(1)	(1)	(2)	CP0	—	—
bit 7							bit 0

Legend:			
R = Readable bit	WO = Write-Once bit	U = Unimplemented bit	t, read as '0'
-n = Value when device is	unprogrammed	'1' = Bit is set	'0' = Bit is cleared

bit 7-3 Unimplemented: Read as '0'

bit 2

CP0: Code Protection bit

- 1 = Program memory is not code-protected
- 0 = Program memory is code-protected
- bit 1-0 Unimplemented: Read as '0'
- **Note 1:** The value of these bits in program memory should always be '1'. This ensures that the location is executed as a NOP if it is accidentally executed.
 - **2:** This bit should always be maintained as '0'.

R/WO-1	R/WO-1	U-0	R/WO-1	R/WO-1	R/WO-1	R/WO-1	R/WO-1
IESO	FCMEN	—	LPT1OSC	T1DIG	FOSC2	FOSC1	FOSC0
bit 7	•	•	-	•	•		bit (
Legend:							
R = Readabl	le bit	WO = Write-	Once bit	U = Unimplen	nented bit, read	l as '0'	
-n = Value w	hen device is ur	programmed		'1' = Bit is set		'0' = Bit is clea	ared
bit 7		ood Start up	(Internal/Externation	al Occillator Su	uitchovor) Cont	rol hit	
	1 = Two-Spee	•			vitchover) Cont		
	0 = Two-Spee						
bit 6	•	•	onitor Enable bi	t			
	1 = Fail-Safe	Clock Monitor	enabled				
	0 = Fail-Safe	Clock Monitor	disabled				
bit 5	Unimplemen	ted: Read as	'0'				
bit 4	LPT1OSC: T	10SC/SOSC I	Power Selection	Configuration	bit		
			SC circuit selec				
	•		SC circuit select				
bit 3		•	out Clock Enabl				
			digital input with is a digital input	•			
bit 2-0	FOSC<2:0>:		•	without chabin	IS TOODEN		
511 2 0			as primary; EC	PLL oscillator	with PLL enabl	ed [.] CLKO on F	8A6
			primary; extern				
			as primary; hig	•		n software PLL	control
			primary; high-sp			4	
			cillator block wit lator block with				
			cillator block with				RA7
	000 = INTIO2	2 internal oscil	lator block with	I/O on RA6 and	d RA7		

REGISTER 25-3: CONFIG2L: CONFIGURATION REGISTER 2 LOW (BYTE ADDRESS 300002h)

REGISTER 25-4: CONFIG2H: CONFIGURATION REGISTER 2 HIGH (BYTE ADDRESS 300003h)

U-0	U-0	U-0	U-0	R/WO-1	R/WO-1	R/WO-1	R/WO-1
(1)	(1)	(1)	(1)	WDTPS3	WDTPS2	WDTPS1	WDTPS0
bit 7	·					•	bit C
Legend:							
R = Reada	ble bit	WO = Write-O	nce bit	U = Unimplem	nented bit, read	l as '0'	
-n = Value	when device is ur	programmed		'1' = Bit is set		'0' = Bit is clea	ared
bit 7-4	Unimplemen	ted: Read as '0	,				
bit 3-0	WDTPS<3:0>	·: Watchdog Tin	ner Postscale	Select bits			
	1111 = 1:32 ,7	768					
	1110 = 1:16 ,3						
	1101 = 1:8,19						
	1100 = 1:4,09						
	1011 = 1:2,04 1010 = 1:1,02						
	1010 = 1.1,02 1001 = 1:512						
	1000 = 1:256						
	0111 = 1:128						
	0110 = 1:64						
	0101 = 1:32						

Note 1: The value of these bits in program memory should always be '1'. This ensures that the location is executed as a NOP if it is accidentally executed.

REGISTER 25-5: CONFIG3L: CONFIGURATION REGISTER 3 LOW (BYTE ADDRESS 300004h)

U-0	U-0	U-0	U-0	U-0	U-0	R/WO-1	U-0
(1)	(1)	(1)	(1)	—	—	RTCOSC	—
bit 7							bit 0

Legend:			
R = Readable bit	WO = Write-Once bit	U = Unimplemented b	bit, read as '0'
-n = Value when device	is unprogrammed	'1' = Bit is set	'0' = Bit is cleared

bit 7-2 Unimplemented: Read as '0'

0100 = 1:16 0011 = 1:8 0010 = 1:4 0001 = 1:2 0000 = 1:1

- bit 1 RTCOSC: RTCC Reference Clock Select bit 1 = RTCC uses T1OSC/T1CKI as the reference clock 0 = RTCC uses INTRC as the reference clock
- bit 0 Unimplemented: Read as '0'
- **Note 1:** The value of these bits in program memory should always be '1'. This ensures that the location is executed as a NOP if it is accidentally executed.

REGISTER 25-6: CONFIG3H: CONFIGURATION REGISTER 3 HIGH (BYTE ADDRESS 300005h)

U-0	U-0	U-0	U-0	U-0	U-0	U-0	R/WO-1	
(1)	(1)	(1)	(1)	—	_	_	CCP2MX	
bit 7							bit C	
Legend:								
R = Reada	ble bit	WO = Write-C	Once bit	U = Unimplem	nented bit, read	l as '0'		
-n = Value	when device i	s unprogrammed		'1' = Bit is set '0' = Bit is cleared				
bit 7-1	Unimplen	nented: Read as '	0'					
bit 0	CCP2MX:	CCP2 MUX bit						

1 = CCP2 is multiplexed with RC1 0 = CCP2 is multiplexed with RE7

REGISTER 25-7: DEVID1: DEVICE ID REGISTER 1 FOR PIC18F87J90 FAMILY DEVICES

R	R	R	R	R	R	R	R
DEV2	DEV1	DEV0	REV4	REV3	REV2	REV1	REV0
bit 7							bit 0

Legend:	
R = Read-only bit	

bit 7-5 **DEV<2:0>:** Device ID bits 101 = PIC18F87J90 100 = PIC18F86J90 001 = PIC18F67J90 000 = PIC18F66J90

bit 4-0 **REV<4:0>:** Revision ID bits These bits are used to indicate the device revision.

REGISTER 25-8: DEVID2: DEVICE ID REGISTER 2 FOR PIC18F87J90 FAMILY DEVICES

R	R	R	R	R	R	R	R
DEV10 ⁽¹⁾	DEV9 ⁽¹⁾	DEV8 ⁽¹⁾	DEV7 ⁽¹⁾	DEV6 ⁽¹⁾	DEV5 ⁽¹⁾	DEV4 ⁽¹⁾	DEV3 ⁽¹⁾
bit 7				•			bit 0
Legend:							

R = Read-only bit

bit 7-0 **DEV<10:3>:** Device ID bits⁽¹⁾

These bits are used with the DEV<2:0> bits in the Device ID Register 1 to identify the part number. 0101 0000 = PIC18F87J90 family devices

Note 1: The values for DEV<10:3> may be shared with other device families. The specific device is always identified by using the entire DEV<10:0> bit sequence.

Note 1: The value of these bits in program memory should always be '1'. This ensures that the location is

executed as a NOP if it is accidentally executed.

25.2 Watchdog Timer (WDT)

For PIC18F87J90 family devices, the WDT is driven by the INTRC oscillator. When the WDT is enabled, the clock source is also enabled. The nominal WDT period is 4 ms and has the same stability as the INTRC oscillator.

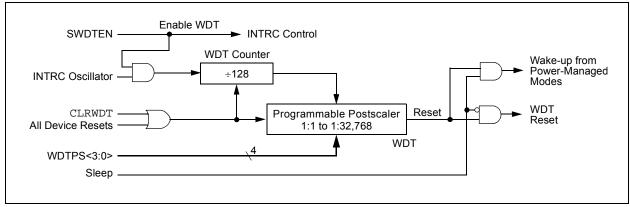
The 4 ms period of the WDT is multiplied by a 16-bit postscaler. Any output of the WDT postscaler is selected by a multiplexor, controlled by the WDTPS bits in Configuration Register 2H. Available periods range from 4 ms to 131.072 seconds (2.18 minutes). The WDT and postscaler are cleared whenever a SLEEP or CLRWDT instruction is executed, or a clock failure (primary or Timer1 oscillator) has occurred.



- Note 1: The CLRWDT and SLEEP instructions clear the WDT and postscaler counts when executed.
 - 2: When a CLRWDT instruction is executed, the postscaler count will be cleared.

25.2.1 CONTROL REGISTER

The WDTCON register (Register 25-9) is a readable and writable register. The SWDTEN bit enables or disables WDT operation. This allows software to override the WDTEN Configuration bit and enable the WDT only if it has been disabled by the Configuration bit.



REGISTER 25-9: WDTCON: WATCHDOG TIMER CONTROL REGISTER

R/W-0	U-0	U-0	U-0	U-0	U-0	U-0	R/W-0				
REGSLP ⁽¹⁾	—	—	_	—	_	– SWDTEN ⁽²⁾					
bit 7											
Legend:											
R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'											
-n = Value at POR '1' = Bit is set				'0' = Bit is cleared x = Bit is unknown							
bit 7	1 = On-chip	Itage Regulator regulator enters	low-power op	eration when d	evice enters SI	eep mode					
0 = On-chip regulator continues to operate normally in Sleep modebit 6-1 Unimplemented: Read as '0'											
bit 0	SWDTEN: Software Controlled Watchdog Timer Enable bit ⁽²⁾ 1 = Watchdog Timer is on										

- 0 = Watchdog Timer is off
- Note 1: The REGSLP bit is automatically cleared when a Low-Voltage Detect condition occurs.
 - **2:** This bit has no effect if the Configuration bit, WDTEN, is enabled.

TABLE 25-3: SUMMARY OF WATCHDOG TIMER REGISTERS

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
RCON	IPEN	_	CM	RI	TO	PD	POR	BOR	60
WDTCON	REGSLP	_	_	_	_	_		SWDTEN	60

Legend: — = unimplemented, read as '0'. Shaded cells are not used by the Watchdog Timer.

25.3 On-Chip Voltage Regulator

All of the PIC18F87J90 family devices power their core digital logic at a nominal 2.5V. For designs that are required to operate at a higher typical voltage, such as 3.3V, all devices in the PIC18F87J90 family incorporate an on-chip regulator that allows the device to run its core logic from VDD.

The regulator is controlled by the ENVREG pin. Tying VDD to the pin enables the regulator, which in turn, provides power to the core from the other VDD pins. When the regulator is enabled, a low-ESR filter capacitor must be connected to the VDDCORE/VCAP pin (Figure 25-2). This helps to maintain the stability of the regulator. The recommended value for the filter capacitor is provided in Section 28.3 "DC Characteristics: PIC18F87J90 Family (Industrial)".

If ENVREG is tied to Vss, the regulator is disabled. In this case, separate power for the core logic at a nominal 2.5V must be supplied to the device on the VDDCORE/VCAP pin to run the I/O pins at higher voltage levels, typically 3.3V. Alternatively, the VDDCORE/VCAP and VDD pins can be tied together to operate at a lower nominal voltage. Refer to Figure 25-2 for possible configurations.

25.3.1 VOLTAGE REGULATION AND LOW-VOLTAGE DETECTION

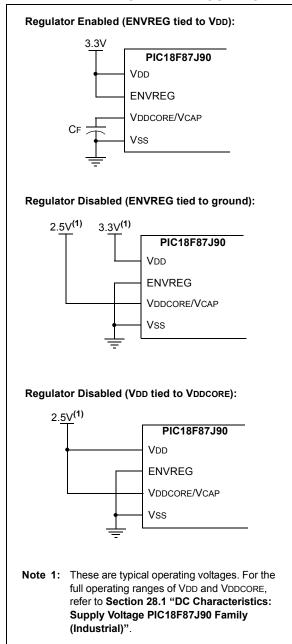
When it is enabled, the on-chip regulator provides a constant voltage of 2.5V nominal to the digital core logic. The regulator can provide this level from a VDD of about 2.5V, all the way up to the device's VDDMAX. It does not have the capability to boost VDD levels below 2.5V.

In order to prevent "brown-out" conditions when the voltage drops too low for the regulator, the regulator enters Tracking mode. In Tracking mode, the regulator output follows VDD with a typical voltage drop of 100 mV.

The on-chip regulator includes a simple Low-Voltage Detect (LVD) circuit. If VDD drops too low to maintain approximately 2.45V on VDDCORE, the circuit sets the Low-Voltage Detect Interrupt Flag, LVDIF (PIR2<2>), and clears the REGSLP (WDTCON<7>) bit, if it was set.

This can be used to generate an interrupt and put the application into a low-power operational mode, or trigger an orderly shutdown. Low-Voltage Detection is only available when the regulator is enabled.

FIGURE 25-2: CONNECTIONS FOR THE ON-CHIP REGULATOR



25.3.2 ON-CHIP REGULATOR AND BOR

When the on-chip regulator is enabled, PIC18F87J90 family devices also have a simple Brown-out Reset capability. If the voltage supplied to the regulator falls to a level that is inadequate to maintain a regulated output for full-speed operation, the regulator Reset circuitry will generate a Brown-out Reset. This event is captured by the BOR flag bit (RCON<0>).

The operation of the BOR is described in more detail in Section 5.4 "Brown-out Reset (BOR)" and Section 5.4.1 "Detecting BOR".

25.3.3 POWER-UP REQUIREMENTS

The on-chip regulator is designed to meet the power-up requirements for the device. If the application does not use the regulator, then strict power-up conditions must be adhered to. While powering up, VDDCORE must never exceed VDD by 0.3 volts.

25.3.4 OPERATION IN SLEEP MODE

When enabled, the on-chip regulator always consumes a small incremental amount of current over IDD. This includes when the device is in Sleep mode, even though the core digital logic does not require power. To provide additional savings in applications where power resources are critical, the regulator can be configured to automatically disable itself whenever the device goes into Sleep mode. This feature is controlled by the REGSLP bit (WDTCON<7>). Setting this bit disables the regulator in Sleep mode, and reduces its current consumption to a minimum.

Substantial Sleep mode power savings can be obtained by setting the REGSLP bit, but device wake-up time will increase in order to ensure the regulator has enough time to stabilize. The REGSLP bit is automatically cleared by hardware when a Low-Voltage Detect condition occurs. The REGSLP bit can be set again in software, which would continue to keep the voltage regulator in Low-Power mode. This, however, is not recommended if any write operations to the Flash will be performed.

25.4 Two-Speed Start-up

The Two-Speed Start-up feature helps to minimize the latency period, from oscillator start-up to code execution, by allowing the microcontroller to use the INTRC oscillator as a clock source until the primary clock source is available. It is enabled by setting the IESO Configuration bit.

Two-Speed Start-up should be enabled only if the Primary Oscillator mode is HS or HSPLL (Crystal-Based) modes. Since the EC and ECPLL modes do not require an OST start-up delay, Two-Speed Start-up should be disabled.

When enabled, Resets and wake-ups from Sleep mode cause the device to configure itself to run from the internal oscillator block as the clock source, following the time-out of the Power-up Timer after a Power-on Reset is enabled. This allows almost immediate code execution while the primary oscillator starts and the OST is running. Once the OST times out, the device automatically switches to PRI_RUN mode.

In all other power-managed modes, Two-Speed Start-up is not used. The device will be clocked by the currently selected clock source until the primary clock source becomes available. The setting of the IESO bit is ignored.

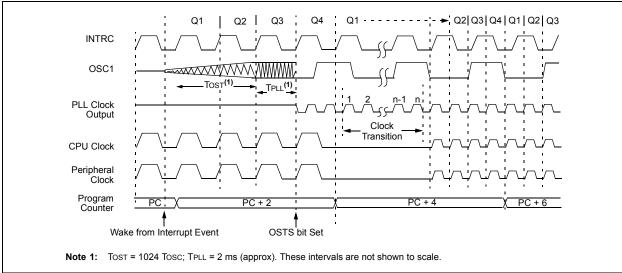


FIGURE 25-3: TIMING TRANSITION FOR TWO-SPEED START-UP (INTRC TO HSPLL)

25.4.1 SPECIAL CONSIDERATIONS FOR USING TWO-SPEED START-UP

While using the INTRC oscillator in Two-Speed Start-up, the device still obeys the normal command sequences for entering power-managed modes, including serial SLEEP instructions (refer to **Section 4.1.4 "Multiple Sleep Commands**"). In practice, this means that user code can change the SCS<1:0> bits setting or issue SLEEP instructions before the OST times out. This would allow an application to briefly wake-up, perform routine "housekeeping" tasks and return to Sleep before the device starts to operate from the primary oscillator.

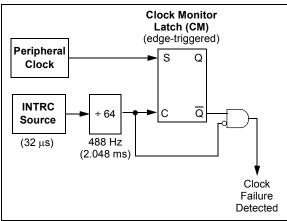
User code can also check if the primary clock source is currently providing the device clocking by checking the status of the OSTS bit (OSCCON<3>). If the bit is set, the primary oscillator is providing the clock. Otherwise, the internal oscillator block is providing the clock during wake-up from Reset or Sleep mode.

25.5 Fail-Safe Clock Monitor

The Fail-Safe Clock Monitor (FSCM) allows the microcontroller to continue operation in the event of an external oscillator failure by automatically switching the device clock to the internal oscillator block. The FSCM function is enabled by setting the FCMEN Configuration bit.

When FSCM is enabled, the INTRC oscillator runs at all times to monitor clocks to peripherals and provides a backup clock in the event of a clock failure. Clock monitoring (shown in Figure 25-4) is accomplished by creating a sample clock signal which is the INTRC output divided by 64. This allows ample time between FSCM sample clocks for a peripheral clock edge to occur. The peripheral device clock and the sample clock are presented as inputs to the Clock Monitor (CM) latch. The CM is set on the falling edge of the device clock source but cleared on the rising edge of the sample clock.

FIGURE 25-4: FSCM BLOCK DIAGRAM



Clock failure is tested for on the falling edge of the sample clock. If a sample clock falling edge occurs while CM is still set, a clock failure has been detected (Figure 25-5). This causes the following:

- the FSCM generates an oscillator fail interrupt by setting bit, OSCFIF (PIR2<7>);
- the device clock source is switched to the internal oscillator block (OSCCON is not updated to show the current clock source – this is the Fail-Safe condition); and
- the WDT is reset.

During switchover, the postscaler frequency from the internal oscillator block may not be sufficiently stable for timing-sensitive applications. In these cases, it may be desirable to select another clock configuration and enter an alternate power-managed mode. This can be done to attempt a partial recovery or execute a controlled shutdown. See Section 4.1.4 "Multiple Sleep Commands" and Section 25.4.1 "Special Considerations for Using Two-Speed Start-up" for more details.

The FSCM will detect failures of the primary or secondary clock sources only. If the internal oscillator block fails, no failure would be detected, nor would any action be possible.

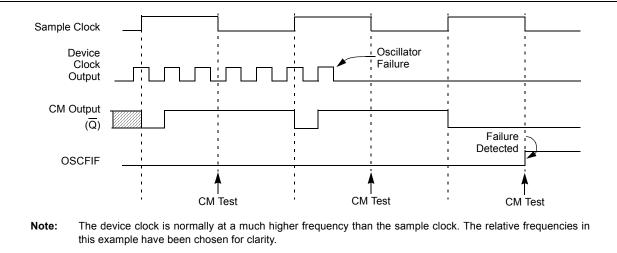
25.5.1 FSCM AND THE WATCHDOG TIMER

Both the FSCM and the WDT are clocked by the INTRC oscillator. Since the WDT operates with a separate divider and counter, disabling the WDT has no effect on the operation of the INTRC oscillator when the FSCM is enabled.

As already noted, the clock source is switched to the INTRC clock when a clock failure is detected; this may mean a substantial change in the speed of code execution. If the WDT is enabled with a small prescale value, a decrease in clock speed allows a WDT time-out to occur and a subsequent device Reset. For this reason, Fail-Safe Clock Monitor events also reset the WDT and postscaler, allowing it to start timing from when execution speed was changed and decreasing the likelihood of an erroneous time-out.

If the interrupt is disabled, subsequent interrupts while in Idle mode will cause the CPU to begin executing instructions while being clocked by the INTRC source.





25.5.2 EXITING FAIL-SAFE OPERATION

The Fail-Safe condition is terminated by either a device Reset or by entering a power-managed mode. On Reset, the controller starts the primary clock source specified in Configuration Register 2H (with any required start-up delays that are required for the oscillator mode, such as the OST or PLL timer). The INTRC oscillator provides the device clock until the primary clock source becomes ready (similar to a Two-Speed Start-up). The clock source is then switched to the primary clock (indicated by the OSTS bit in the OSCCON register becoming set). The Fail-Safe Clock Monitor then resumes monitoring the peripheral clock.

The primary clock source may never become ready during start-up. In this case, operation is clocked by the INTOSC multiplexor. The OSCCON register will remain in its Reset state until a power-managed mode is entered.

25.5.3 FSCM INTERRUPTS IN POWER-MANAGED MODES

By entering a power-managed mode, the clock multiplexor selects the clock source selected by the OSCCON register. Fail-Safe Clock Monitoring of the power-managed clock source resumes in the power-managed mode.

If an oscillator failure occurs during power-managed operation, the subsequent events depend on whether or not the oscillator failure interrupt is enabled. If enabled (OSCFIF = 1), code execution will be clocked by the INTRC multiplexor. An automatic transition back to the failed clock source will not occur.

25.5.4 POR OR WAKE-UP FROM SLEEP

The FSCM is designed to detect oscillator failure at any point after the device has exited Power-on Reset (POR) or low-power Sleep mode. When the primary device clock is either EC or INTRC mode, monitoring can begin immediately following these events.

For HS or HSPLL modes, the situation is somewhat different. Since the oscillator may require a start-up time considerably longer than the FSCM sample clock time, a false clock failure may be detected. To prevent this, the internal oscillator block is automatically configured as the device clock and functions until the primary clock is stable (the OST and PLL timers have timed out). This is identical to Two-Speed Start-up mode. Once the primary clock is stable, the INTRC returns to its role as the FSCM source.

Note: The same logic that prevents false oscillator failure interrupts on POR, or wake from Sleep, will also prevent the detection of the oscillator's failure to start at all following these events. This can be avoided by monitoring the OSTS bit and using a timing routine to determine if the oscillator is taking too long to start. Even so, no oscillator failure interrupt will be flagged.

As noted in Section 25.4.1 "Special Considerations for Using Two-Speed Start-up", it is also possible to select another clock configuration and enter an alternate power-managed mode while waiting for the primary clock to become stable. When the new power-managed mode is selected, the primary clock is disabled.

25.6 Program Verification and Code Protection

For all devices in the PIC18F87J90 family of devices, the on-chip program memory space is treated as a single block. Code protection for this block is controlled by one Configuration bit, CP0. This bit inhibits external reads and writes to the program memory space. It has no direct effect in normal execution mode.

25.6.1 CONFIGURATION REGISTER PROTECTION

The Configuration registers are protected against untoward changes or reads in two ways. The primary protection is the write-once feature of the Configuration bits which prevents reconfiguration once the bit has been programmed during a power cycle. To safeguard against unpredictable events, Configuration bit changes resulting from individual cell-level disruptions (such as ESD events) will cause a parity error and trigger a device Reset.

The data for the Configuration registers is derived from the Flash Configuration Words in program memory. When the CP0 bit set, the source data for device configuration is also protected as a consequence.

25.7 In-Circuit Serial Programming

PIC18F87J90 family microcontrollers can be serially programmed while in the end application circuit. This is simply done with two lines for clock and data, and three other lines for power, ground and the programming voltage. This allows customers to manufacture boards with unprogrammed devices and then program the microcontroller just before shipping the product. This also allows the most recent firmware or a custom firmware to be programmed.

25.8 In-Circuit Debugger

When the DEBUG Configuration bit is programmed to a '0', the In-Circuit Debugger functionality is enabled. This function allows simple debugging functions when used with MPLAB[®] IDE. When the microcontroller has this feature enabled, some resources are not available for general use. Table 25-4 shows which resources are required by the background debugger.

TABLE 25-4: DEBUGGER RESOURCES

I/O pins:	RB6, RB7
Stack:	2 levels
Program Memory:	512 bytes
Data Memory:	10 bytes

NOTES:

26.0 INSTRUCTION SET SUMMARY

The PIC18F87J90 family of devices incorporate the standard set of 75 PIC18 core instructions, as well as an extended set of 8 new instructions for the optimization of code that is recursive or that utilizes a software stack. The extended set is discussed later in this section.

26.1 Standard Instruction Set

The standard PIC18 MCU instruction set adds many enhancements to the previous PIC[®] MCU instruction sets, while maintaining an easy migration from these PIC MCU instruction sets. Most instructions are a single program memory word (16 bits), but there are four instructions that require two program memory locations.

Each single-word instruction is a 16-bit word divided into an opcode, which specifies the instruction type and one or more operands, which further specify the operation of the instruction.

The instruction set is highly orthogonal and is grouped into four basic categories:

- Byte-oriented operations
- Bit-oriented operations
- · Literal operations
- Control operations

The PIC18 instruction set summary in Table 26-2 lists **byte-oriented**, **bit-oriented**, **literal** and **control** operations. Table 26-1 shows the opcode field descriptions.

Most byte-oriented instructions have three operands:

- 1. The file register (specified by 'f')
- 2. The destination of the result (specified by 'd')
- 3. The accessed memory (specified by 'a')

The file register designator, 'f', specifies which file register is to be used by the instruction. The destination designator, 'd', specifies where the result of the operation is to be placed. If 'd' is zero, the result is placed in the WREG register. If 'd' is one, the result is placed in the file register specified in the instruction.

All **bit-oriented** instructions have three operands:

- 1. The file register (specified by 'f')
- 2. The bit in the file register (specified by 'b')
- 3. The accessed memory (specified by 'a')

The bit field designator, 'b', selects the number of the bit affected by the operation, while the file register designator, 'f', represents the number of the file in which the bit is located.

The **literal** instructions may use some of the following operands:

- A literal value to be loaded into a file register (specified by 'k')
- The desired FSR register to load the literal value into (specified by 'f')
- No operand required (specified by '—')

The **control** instructions may use some of the following operands:

- A program memory address (specified by 'n')
- The mode of the CALL or RETURN instructions (specified by 's')
- The mode of the table read and table write instructions (specified by 'm')
- No operand required (specified by '—')

All instructions are a single word, except for four double-word instructions. These instructions were made double-word to contain the required information in 32 bits. In the second word, the 4 MSbs are '1's. If this second word is executed as an instruction (by itself), it will execute as a NOP.

All single-word instructions are executed in a single instruction cycle, unless a conditional test is true or the program counter is changed as a result of the instruction. In these cases, the execution takes two instruction cycles with the additional instruction cycle(s) executed as a NOP.

The double-word instructions execute in two instruction cycles.

One instruction cycle consists of four oscillator periods. Thus, for an oscillator frequency of 4 MHz, the normal instruction execution time is 1 μ s. If a conditional test is true, or the program counter is changed as a result of an instruction, the instruction execution time is 2 μ s. Two-word branch instructions (if true) would take 3 μ s.

Figure 26-1 shows the general formats that the instructions can have. All examples use the convention 'nnh' to represent a hexadecimal number.

The Instruction Set Summary, shown in Table 26-2, lists the standard instructions recognized by the Microchip MPASM[™] Assembler.

Section 26.1.1 "Standard Instruction Set" provides a description of each instruction.

TABLE 26-1: OPCODE FIELD DESCRIPTIONS

Field	Description
a	RAM access bit:
	a = 0: RAM location in Access RAM (BSR register is ignored)
	a = 1: RAM bank is specified by BSR register
bbb	Bit address within an 8-bit file register (0 to 7).
BSR	Bank Select Register. Used to select the current RAM bank.
C, DC, Z, OV, N	ALU Status bits: Carry, Digit Carry, Zero, Overflow, Negative.
d	Destination select bit:
	d = 0: store result in WREG
	d = 1: store result in file register f
dest	Destination: either the WREG register or the specified register file location.
f	8-bit Register file address (00h to FFh), or 2-bit FSR designator (0h to 3h).
fs	12-bit Register file address (000h to FFFh). This is the source address.
f _d	12-bit Register file address (000h to FFFh). This is the destination address.
GIE	Global Interrupt Enable bit.
k	Literal field, constant data or label (may be either an 8-bit, 12-bit or a 20-bit value).
label	Label name.
mm	The mode of the TBLPTR register for the table read and table write instructions. Only used with table read and table write instructions:
*	No Change to register (such as TBLPTR with table reads and writes)
*+	Post-Increment register (such as TBLPTR with table reads and writes)
*_	Post-Decrement register (such as TBLPTR with table reads and writes)
+*	Pre-Increment register (such as TBLPTR with table reads and writes)
n	The relative address (2's complement number) for relative branch instructions or the direct address for
	Call/Branch and Return instructions.
PC	Program Counter.
PCL	Program Counter Low Byte.
PCH	Program Counter High Byte.
PCLATH	Program Counter High Byte Latch.
PCLATU	Program Counter Upper Byte Latch.
PD	Power-Down bit.
PRODH	Product of Multiply High Byte.
PRODL	Product of Multiply Low Byte.
S	Fast Call/Return mode select bit:
	s = 0: do not update into/from shadow registers s = 1: certain registers loaded into/from shadow registers (Fast mode)
TBLPTR	21-bit Table Pointer (points to a Program Memory location).
TABLAT	8-bit Table Latch.
TO	Time-out bit.
TOS	Top-of-Stack.
u	Unused or Unchanged.
WDT	Watchdog Timer.
WREG	Working register (accumulator).
x	Don't care ('0' or '1'). The assembler will generate code with $x = 0$. It is the recommended form of use for
	compatibility with all Microchip software tools.
z _s	7-bit offset value for Indirect Addressing of register files (source).
z _d	7-bit offset value for Indirect Addressing of register files (destination).
{ }	Optional argument.
[text]	Indicates an Indexed Address.
(text)	The contents of text.
[expr] <n></n>	Specifies bit n of the register indicated by the pointer expr.
\rightarrow	Assigned to.
< >	Register bit field.
E	In the set of.
italics	User-defined term (font is Courier New).

Byte-oriented file register operations	Example Instruction
15 10 9 8 7 0	
OPCODE d a f (FILE #)	ADDWF MYREG, W, B
 d = 0 for result destination to be WREG register d = 1 for result destination to be file register (f) a = 0 to force Access Bank a = 1 for BSR to select bank f = 8-bit file register address 	
Byte to Byte move operations (2-word)	
<u>15 12 11 0</u>	
OPCODE f (Source FILE #)	MOVFF MYREG1, MYREG2
15 12 11 0	
1111 f (Destination FILE #)	
f = 12-bit file register address	
Bit-oriented file register operations	
<u>15 12 11 9 8 7 0</u>	
OPCODE b (BIT #) a f (FILE #)	BSF MYREG, bit, B
 b = 3-bit position of bit in file register (f) a = 0 to force Access Bank a = 1 for BSR to select bank f = 8-bit file register address 	
Literal operations	
15 8 7 0	
OPCODE k (literal)	MOVLW 7Fh
k = 8-bit immediate value	
Control operations	
CALL, GOTO and Branch operations	
15 8 7 0	
OPCODE n<7:0> (literal)	GOTO Label
15 12 11 0	
1111 n<19:8> (literal)	
n = 20-bit immediate value	
15 8 7 0	
OPCODE S n<7:0> (literal)	CALL MYFUNC
15 12 11 0	
1111 n<19:8> (literal)	
S = Fast bit	
15 11 10 0	
OPCODE n<10:0> (literal)	BRA MYFUNC
15 8 7 0	
OPCODE n<7:0> (literal)	BC MYFUNC

TABLE 26-2: PIC18F87J90 FAMILY INSTRUCTION SET

Mnemo	onic,	Description	Gualaa	16-E	Bit Instr	uction V	Vord	Status	Natas
Opera	nds	Description	Cycles	MSb			LSb	Affected	Notes
BYTE-ORI	ENTED (OPERATIONS							
ADDWF	f, d, a	Add WREG and f	1	0010	01da	ffff	ffff	C, DC, Z, OV, N	1, 2
ADDWFC	f, d, a	Add WREG and Carry bit to f	1	0010	00da	ffff	ffff	C, DC, Z, OV, N	
ANDWF	f, d, a	AND WREG with f	1	0001	01da	ffff	ffff	Z, N	1,2
CLRF	f, a	Clear f	1	0110	101a	ffff	ffff	Z	2
COMF	f, d, a	Complement f	1	0001	11da	ffff	ffff	Z, N	1, 2
CPFSEQ	f, a	Compare f with WREG, Skip =	1 (2 or 3)	0110	001a	ffff	ffff	None	4
CPFSGT	f, a	Compare f with WREG, Skip >	1 (2 or 3)	0110	010a	ffff	ffff	None	4
CPFSLT	f, a	Compare f with WREG, Skip <	1 (2 or 3)	0110	000a	ffff	ffff	None	1, 2
DECF	f, d, a	Decrement f	1	0000	01da	ffff	ffff	C, DC, Z, OV, N	1, 2, 3, 4
DECFSZ	f, d, a	Decrement f, Skip if 0	1 (2 or 3)	0010	11da	ffff	ffff	None	1, 2, 3, 4
DCFSNZ	f, d, a	Decrement f, Skip if Not 0	1 (2 or 3)	0100	11da	ffff	ffff	None	1, 2
INCF	f, d, a	Increment f	1 ΄	0010	10da	ffff	ffff	C, DC, Z, OV, N	1, 2, 3, 4
INCFSZ	f, d, a	Increment f, Skip if 0	1 (2 or 3)	0011	11da	ffff	ffff	None	4
INFSNZ	f, d, a	Increment f, Skip if Not 0	1 (2 or 3)	0100	10da	ffff	ffff	None	1, 2
IORWF	f, d, a	Inclusive OR WREG with f	1 ΄	0001	00da	ffff	ffff	Z, N	1, 2
MOVF	f, d, a	Move f	1	0101	00da	ffff	ffff	Z, N	1
MOVFF	f _s , f _d	Move f _s (source) to 1st word	2	1100	ffff	ffff		None	
	3, u	f _d (destination) 2nd word		1111	ffff	ffff	ffff		
MOVWF	f, a	Move WREG to f	1	0110	111a	ffff	ffff	None	
MULWF	f, a	Multiply WREG with f	1	0000	001a	ffff	ffff	None	1, 2
NEGF	f, a	Negate f	1	0110	110a	ffff	ffff	C, DC, Z, OV, N	
RLCF	f, d, a	Rotate Left f through Carry	1	0011	01da	ffff	ffff	C, Z, N	1, 2
RLNCF	f, d, a	Rotate Left f (No Carry)	1	0100	01da	ffff	ffff	Z, N	
RRCF	f, d, a	Rotate Right f through Carry	1	0011	00da	ffff	ffff	C, Z, N	
RRNCF	f, d, a	Rotate Right f (No Carry)	1	0100	00da	ffff	ffff		
SETF	f, a	Set f	1	0110	100a	ffff	ffff	None	1, 2
SUBFWB	f, d, a	Subtract f from WREG with	1	0101	01da	ffff	ffff	C, DC, Z, OV, N	,
		Borrow							
SUBWF	f, d, a	Subtract WREG from f	1	0101	11da	ffff	ffff	C, DC, Z, OV, N	1, 2
SUBWFB	f, d, a	Subtract WREG from f with	1		10da	ffff	ffff	C, DC, Z, OV, N	
	, , -	Borrow						. , , ,	
SWAPF	f, d, a	Swap Nibbles in f	1	0011	10da	ffff	ffff	None	4
TSTFSZ	f, a	Test f, Skip if 0	1 (2 or 3)	0110	011a	ffff	ffff	None	1, 2
XORWF	f, d, a	Exclusive OR WREG with f	1		10da	ffff		Z, N	,

Note 1: When a PORT register is modified as a function of itself (e.g., MOVF PORTB, 1, 0), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and, where applicable, d = 1), the prescaler will be cleared if assigned.

3: If Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

4: Some instructions are two-word instructions. The second word of these instructions will be executed as a NOP unless the first word of the instruction retrieves the information embedded in these 16 bits. This ensures that all program memory locations have a valid instruction.

Mnem	onic,	Description	Ouslas	16-E	Bit Instr	uction V	Vord	Status	Nata -
Opera		Description	Cycles	MSb			LSb	Affected	Notes
BIT-ORIEN	NTED OP	ERATIONS							
BCF	f, b, a	Bit Clear f	1	1001	bbba	ffff	ffff	None	1, 2
BSF	f, b, a	Bit Set f	1	1000	bbba	ffff	ffff	None	1, 2
BTFSC	f, b, a	Bit Test f, Skip if Clear	1 (2 or 3)	1011	bbba	ffff	ffff	None	3, 4
BTFSS	f, b, a	Bit Test f, Skip if Set	1 (2 or 3)	1010	bbba	ffff	ffff	None	3, 4
BTG	f, b, a	Bit Toggle f	1	0111	bbba	ffff	ffff	None	1, 2
CONTROL	OPERA	TIONS							
BC	n	Branch if Carry	1 (2)	1110	0010	nnnn	nnnn	None	
BN	n	Branch if Negative	1 (2)	1110	0110	nnnn	nnnn	None	
BNC	n	Branch if Not Carry	1 (2)	1110	0011	nnnn	nnnn	None	
BNN	n	Branch if Not Negative	1 (2)	1110	0111	nnnn	nnnn	None	
BNOV	n	Branch if Not Overflow	1 (2)	1110	0101	nnnn	nnnn	None	
BNZ	n	Branch if Not Zero	1 (2)	1110	0001	nnnn	nnnn	None	
BOV	n	Branch if Overflow	1 (2)	1110	0100	nnnn	nnnn	None	
BRA	n	Branch Unconditionally	2	1101	0nnn	nnnn		None	
BZ	n	Branch if Zero	1 (2)	1110	0000	nnnn	nnnn	None	
CALL	n, s	Call Subroutine 1st word	2	1110	110s	kkkk	kkkk	None	
		2nd word		1111	kkkk	kkkk	kkkk		
CLRWDT		Clear Watchdog Timer	1	0000	0000	0000	0100	TO, PD	
DAW	—	Decimal Adjust WREG	1	0000	0000	0000	0111	С	
GOTO	n	Go to Address 1st word	2	1110	1111	kkkk	kkkk	None	
		2nd word		1111	kkkk	kkkk	kkkk		
NOP	—	No Operation	1	0000	0000	0000	0000	None	
NOP	—	No Operation	1	1111		XXXX	xxxx	None	4
POP	—	Pop Top of Return Stack (TOS)	1	0000	0000	0000	0110	None	
PUSH	—	Push Top of Return Stack (TOS)		0000	0000	0000	0101	None	
RCALL	n	Relative Call	2	1101	1nnn	nnnn		None	
RESET		Software Device Reset	1	0000	0000	1111		All	
RETFIE	S	Return from Interrupt Enable	2	0000	0000	0001	000s	GIE/GIEH, PEIE/GIEL	
RETLW	k	Return with Literal in WREG	2	0000	1100	kkkk	kkkk	None	
RETURN	S	Return from Subroutine	2	0000	0000	0001	001s	None	
SLEEP	_	Go into Standby mode	1	0000	0000	0000	0011	TO, PD	

TABLE 26-2:	PIC18F87J90 FAMILY INSTRUCTION SET ((CONTINUED))
			<i>i</i>

Note 1: When a PORT register is modified as a function of itself (e.g., MOVF PORTB, 1, 0), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and, where applicable, d = 1), the prescaler will be cleared if assigned.

3: If Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

4: Some instructions are two-word instructions. The second word of these instructions will be executed as a NOP unless the first word of the instruction retrieves the information embedded in these 16 bits. This ensures that all program memory locations have a valid instruction.

TABLE 20						/				
Mnem	-	Description	Cycles	16-Bit Instruction Word				Status	Notes	
Operands		Decemption	Cycloc	MSb			LSb	Affected		
LITERAL (OPERA	TIONS								
ADDLW	k	Add Literal and WREG	1	0000	1111	kkkk	kkkk	C, DC, Z, OV, N		
ANDLW	k	AND Literal with WREG	1	0000	1011	kkkk	kkkk	Z, N		
IORLW	k	Inclusive OR Literal with WREG	1	0000	1001	kkkk	kkkk	Z, N		
LFSR	f, k	Move literal (12-bit) 2nd word	2	1110	1110	00ff	kkkk	None		
		to FSR(f) 1st word		1111	0000	kkkk	kkkk			
MOVLB	k	Move Literal to BSR<3:0>	1	0000	0001	0000	kkkk	None		
MOVLW	k	Move Literal to WREG	1	0000	1110	kkkk	kkkk	None		
MULLW	k	Multiply Literal with WREG	1	0000	1101	kkkk	kkkk	None		
RETLW	k	Return with Literal in WREG	2	0000	1100	kkkk	kkkk	None		
SUBLW	k	Subtract WREG from Literal	1	0000	1000	kkkk	kkkk	C, DC, Z, OV, N		
XORLW	k	Exclusive OR Literal with WREG	1	0000	1010	kkkk	kkkk	Z, N		
DATA MEN	NORY +	PROGRAM MEMORY OPERATI	ONS							
TBLRD*		Table Read	2	0000	0000	0000	1000	None		
TBLRD*+		Table Read with Post-Increment		0000	0000	0000	1001	None		
TBLRD*-		Table Read with Post-Decrement		0000	0000	0000	1010	None		
TBLRD+*		Table Read with Pre-Increment		0000	0000	0000	1011	None		
TBLWT*		Table Write	2	0000	0000	0000	1100	None		
TBLWT*+		Table Write with Post-Increment		0000	0000	0000	1101	None		
TBLWT*-		Table Write with Post-Decrement		0000	0000	0000	1110	None		
TBLWT+*		Table Write with Pre-Increment		0000	0000	0000	1111	None		

TABLE 26-2: PIC18F87J90 FAMILY INSTRUCTION SET (CONTINUED)

Note 1: When a PORT register is modified as a function of itself (e.g., MOVF PORTB, 1, 0), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and, where applicable, d = 1), the prescaler will be cleared if assigned.

3: If Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

4: Some instructions are two-word instructions. The second word of these instructions will be executed as a NOP unless the first word of the instruction retrieves the information embedded in these 16 bits. This ensures that all program memory locations have a valid instruction.

ADD W to f

26.1.1 STANDARD INSTRUCTION SET

ADDLW	ADD Litera	al to W		ADDWF
Syntax:	ADDLW	k		Syntax:
Operands:	$0 \le k \le 255$			Operands:
Operation:	$(W) + k \rightarrow$	W		
Status Affected:	N, OV, C, [DC, Z		Operation:
Encoding:	0000	1111 kk	kk kkkk	Status Affected:
Description:		its of W are ad 'k' and the res	dded to the sult is placed in	Encoding: Description:
Words:	1			
Cycles:	1			
Q Cycle Activity:				
Q1	Q2	Q3	Q4	
Decode	Read literal 'k'	Process Data	Write to W	
Example: Before Instru W =		L5h		
After Instruct				Words:
W =	25h			
				Cycles: Q Cycle Activit
				Q Cycle Activit Q1
				Decode
				Example:
				Before Ins: W REG After Instru W REG

ADDWF f {,d {,a}} $0 \leq f \leq 255$ $d \in [0,1]$ a ∈ [0,1] $(W) + (f) \rightarrow dest$ N, OV, C, DC, Z d: 0010 01da ffff ffff Add W to register 'f'. If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'. If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank. If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \le 95$ (5Fh). See Section 26.2.3 "Byte-Oriented and **Bit-Oriented Instructions in Indexed** Literal Offset Mode" for details. 1 1 ity: Q4 Q3 Q2 е Read Process Write to register 'f Data destination ADDWF REG, 0, 0 struction = 17h = 0C2h ruction 0D9h = REG = 0C2h

Note: All PIC18 instructions may take an optional label argument preceding the instruction mnemonic for use in symbolic addressing. If a label is used, the instruction format then becomes: {label} instruction argument(s).

ADD W and	d Carry I	bit to	f			
ADDWFC	f {,d {,	a}}				
$0 \le f \le 255$ d $\in [0,1]$ a $\in [0,1]$						
(W) + (f) +	$(C) \rightarrow de$	st				
N,OV, C, D	N,OV, C, DC, Z					
0010	0010 00da ffff ffff					
Add W, the Carry flag and data memor location 'f'. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed in data memory location 'f'.						
If 'a' is '0', the Access Bank is select If 'a' is '1', the BSR is used to select t GPR bank.						
If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode," for details						
1						
1						
Q2	Q3			Q4		
Read register 'f'				/rite to stination		
ADDWFC	REG,	0, 3	1			
tion = 1 = 02h = 4Dh on = 0 = 02h						
	ADDWFC $0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$ (W) + (f) + f N,OV, C, D 0010 Add W, the location 'f'. placed in W placed in W mode where Section 26 Bit-Oriente Literal Offs 1 1 2 MDDWFC tion = 1 = 02h = 4Dh on = 0	ADDWFC f {,d {,: $0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$ $a \in [0,1]$ (W) + (f) + (C) \rightarrow devices N,OV, C, DC, Z 0010 00da Add W, the Carry flat location 'f'. If 'd' is '0' placed in W. If 'd' is '0' placed in W. If 'd' is '0' placed in W. If 'd' is '0' placed in data mem If 'a' is '0', the Access If 'a' is '0', the Access If 'a' is '0' and the ex- set is enabled, this i in Indexed Literal O' mode whenever f \le Section 26.2.3 "By Bit-Oriented Instru- Literal Offset Mode 1 1 Q2 Q3 Read Proce register 'f' Data ADDWFC REG, tion = 1 = 02h = 4Dh on = 0	ADDWFC $f \{, d \{, a\}\}$ $0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$ $(W) + (f) + (C) \rightarrow dest$ N,OV, C, DC, Z 0010 00da fff Add W, the Carry flag and location 'f'. If 'd' is '0', the placed in W. If 'd' is '1', th placed in data memory lo If 'a' is '0', the Access Bar If 'a' is '0', the Access Bar If 'a' is '0', the Ascess Bar If 'a' is '0', the Ascess Bar If 'a' is '0', the Ascess Bar If 'a' is '0' and the extended set is enabled, this instruct in Indexed Literal Offset A mode whenever $f \le 95$ (51 Section 26.2.3 "Byte-Ori Bit-Oriented Instruction Literal Offset Mode" for 1 1 2 Q2 Q3 Read register 'f' Data ADDWFC REG, 0, 1 tion = 1 = 02h = 4Dh on = 0	$\begin{array}{llllllllllllllllllllllllllllllllllll$		

ANDLW	AND Litera	al with W	1		
Syntax:	ANDLW	k			
Operands:	$0 \le k \le 255$				
Operation:	(W) .AND.	$k \rightarrow W$			
Status Affected:	N, Z				
Encoding:	0000	1011	kkkk	kkkk	
Description:		e contents of W are ANDed with th it literal 'k'. The result is placed in			
Words:	1				
Cycles:	1				
Q Cycle Activity:					
Q1	Q2	Q3	8	Q4	
Decode	Read literal 'k'	Proce Data		Write to W	
Example: Before Instruc	ANDLW	05Fh			
After Instruction	= A3h				
W	= 03h				

ANDWF	AND W wit	h f					
Syntax:	ANDWF	f {,d {,a}}					
Operands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$						
Operation:	(W) .AND. ((f) \rightarrow dest					
Status Affected:	N, Z						
Encoding:	0001	01da ffi	ff ffff				
Description:	register 'f'. I in W. If 'd' is	The contents of W are ANDed with register 'f'. If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'.					
		If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.					
	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.						
Words:	1						
Cycles:	1						
Q Cycle Activity:							
Q1	Q2	Q3	Q4				
Decode	Read register 'f'	Process Data	Write to destination				
Example:	ANDWF	REG, 0, 0					
Before Instruc	tion						
W REG After Instructio	= 17h = C2h						
W REG	= 02h = C2h						

вс		Branch if (Carry					
Synta	ax:	BC n						
Oper	Operands: $-128 \le n \le 127$							
Oper	ation:	if Carry bit (PC) + 2 +						
Statu	s Affected:	None						
Enco	ding:	1110	0010 nr	nnn nnnn				
Description:			If the Carry bit is '1', then the program will branch.					
		The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a two-cycle instruction.						
Words: 1								
Cycle	es:	1(2)	1(2)					
Q Cy If Ju	ycle Activity: mp:							
	Q1	Q2	Q3	Q4				
	Decode	Read literal 'n'	Process Data	Write to PC				
	No operation	No operation	No operation	No operation				
lf No	o Jump:			•				
	Q1	Q2	Q3	Q4				
	Decode	Read literal 'n'	Process Data	No operation				
<u>Exam</u>	<u>nple:</u>	HERE	BC 5					
	Before Instruc PC After Instructic	= ad	dress (HERI	E)				
	If Carry PC If Carry	= 1; = ad = 0;		E + 12)				
	PC	= ad	dress (HERI	E + 2)				

BCF	Bit Clear f	BN	Branch if N	legative	
Syntax:	BCF f, b {,a}	Syntax:	BN n		
Operands:	$0 \leq f \leq 255$	Operands:	-128 ≤ n ≤ 1	127	
	0 ≤ b ≤ 7 a ∈ [0,1]	Operation:	if Negative (PC) + 2 + 2		
Operation:	$0 \rightarrow f \le b >$	Status Affected:	None		
Status Affected:	None	Encoding:	1110	0110 nni	nn nnnn
Encoding:	1001 bbba ffff ffff	Description:	-	tive bit is '1', th	
Description:	Bit 'b' in register 'f' is cleared.	Docomption	program wi		
	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.		added to the incremente	d to fetch the r	e PC will have next
	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing		,	the new addre n. This instruct istruction.	
	mode whenever f \leq 95 (5Fh). See	Words:	1		
	Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed	Cycles:	1(2)		
	Literal Offset Mode" for details.	Q Cycle Activity:			
Words:	1	If Jump:			
Cycles:	1	Q1	Q2	Q3	Q4
Q Cycle Activity:		Decode	Read literal 'n'	Process Data	Write to PC
Q1	Q2 Q3 Q4	No	No	No	No
Decode	Read Process Write	operation	operation	operation	operation
	register 'f' Data register 'f'	If No Jump:			
		Q1	Q2	Q3	Q4
Example:	BCF FLAG_REG, 7, 0	Decode	Read literal 'n'	Process Data	No operation
Before Instruc				Data	operation
After Instructio	EG = C7h	Example:	HERE	BN Jump	
	EG = 47h	Before Instruc			
		PC After Instruction	= ad	dress (HERE)	
		If Negati PC		dress (Jump)	
		If Negati	ve = 0;	. 1	
		PC	= ad	dress (HERE	+ 2)

Operands: -1 Operation: if (P Status Affected: Ne Encoding: Description: If wi That in: P(tw Words: 1	the Carry to ill branch. he 2's comp dded to the incremented instruction, the C + 2 + 2n.	0011 nmm oit is '0', then plement numb	the program ber '2n' is e PC will have	Syntax Opera Opera Status Encod Descri	nds: tion: Affected: ing:	BNN n $-128 \le n \le$ if Negative (PC) + 2 + None 1110 If the Nega program with	bit is '0', $2n \rightarrow PC$ 0111 n tive bit is '0',		nnnn
Operation: if (P Status Affected: No Encoding: Description: If Th ac inv inv Words: 1	Carry bit is PC) + 2 + 2 lone 1110 the Carry b rill branch. he 2's comp dded to the incremented instruction, the PC + 2 + 2n.	(0, 0) $n \rightarrow PC$ (0011 nnr) (0011 nnr) (0, 0) (0, 0)	the program ber '2n' is e PC will have	Opera Status Encod	tion: Affected: ing:	if Negative (PC) + 2 + None 1110 If the Nega	bit is '0', $2n \rightarrow PC$ 0111 n tive bit is '0',		
(P Status Affected: No Encoding: [Description: If wi Th ac int P(tw Words: 1	PC) + 2 + 2 lone 1110 the Carry b ill branch. he 2's com dded to the incremented instruction, th C + 2 + 2n.	$n \rightarrow PC$ 0011 nnm pit is '0', then plement numb PC. Since the to fetch the n	the program ber '2n' is e PC will have	Status Encod	Affected: ing:	(PC) + 2 + None 1110 If the Nega	$2n \rightarrow PC$ 0111 n tive bit is '0',		
Encoding: Description: If wi Th action in: P(tw Words: 1	1110 i the Carry k vill branch. ihe 2's com dded to the acremented astruction, the C + 2 + 2n.	plement numb PC. Since the to fetch the n	the program ber '2n' is e PC will have	Encod	ing:	1110 If the Nega	tive bit is '0',		
Description: If wi Th ac inv inv Po tw Words: 1	the Carry to ill branch. the 2's comp dded to the incremented instruction, the to + 2 + 2n.	plement numb PC. Since the to fetch the n	the program ber '2n' is e PC will have		0	If the Nega	tive bit is '0',		
wi Th ac in in PC tw Words: 1	vill branch. he 2's comp dded to the incremented instruction, the PC + 2 + 2n.	plement numb PC. Since the to fetch the n	ber '2n' is e PC will have	Descri	ption:	•		then the	ì
ac in in: P(tw Words: 1	dded to the acremented astruction, tl 2C + 2 + 2n.	PC. Since the to fetch the n	e PC will have						
	vo-cycle ins	. This instructi struction.				added to th incremente instruction,	nplement nu e PC. Since d to fetch the the new add n. This instru nstruction.	the PC w e next ress will	vill have I be
Cycles: 1(Words	:	1			
	(2)			Cycles	:	1(2)			
Q Cycle Activity: If Jump:				Q Cyo If Jur	cle Activity: np:				
Q1	Q2	Q3	Q4		Q1	Q2	Q3	(Q4
Decode Rea	ad literal 'n'	Process Data	Write to PC		Decode	Read literal 'n'	Process Data		ite to PC
No	No	No	No		No	No	No	М	١o
operation op	peration	operation	operation		operation	operation	operation	oper	ration
If No Jump:				If No .	Jump:				
Q1	Q2	Q3	Q4	г	Q1	Q2	Q3	1	Q4
Decode Rea	ad literal 'n'	Process Data	No operation		Decode	Read literal 'n'	Process Data		No ration
Example: HI	ERE	BNC Jump		Examp	ble:	HERE	BNN Jum	p	
Before Instruction				В	efore Instruc	tion			
PC	= add	ress (HERE))		PC		dress (HER	E)	
After Instruction				A	fter Instructio				
If Carry PC	= 0; = add	ress (Jump)			If Negati PC		dress (Jum	n)	
If Carry PC	= 1;				If Negati PC			E /	

BNO	v	Branch if N	lot Overflow			в	IZ
Synta	ax:	BNOV n				Sy	ntax:
Oper	ands:	-128 ≤ n ≤ 1	127			Op	erands:
Oper	ation:	if Overflow (PC) + 2 + 2	,			Op	eration:
Statu	s Affected:	None				Sta	atus Affecte
Enco	oding:	1110	0101 nn	nn	nnnn	En	coding:
Desc	cription: If the Overflow bit is '0', then the program will branch.			De	scription:		
		added to the incremented instruction,	nplement num e PC. Since th d to fetch the i the new addre n. This instruct instruction.	e PC next ess w	will have vill be		
Word	ls:	1				We	ords:
Cycle	es:	1(2)				Су	cles:
Q C If Ju	ycle Activity: imp:						Cycle Activ Jump:
	Q1	Q2	Q3		Q4	_	Q1
	Decode	Read literal 'n'	Process Data	W	/rite to PC		Deco
	No operation	No operation	No operation	ор	No eration		No operati
lf No	o Jump:					lf	No Jump:
	Q1	Q2	Q3		Q4	1	Q1
	Decode	Read literal 'n'	Process Data	ор	No eration		Deco
Exan	nple:	HERE	BNOV Jump			Ex	ample:
	Before Instruc PC After Instructio If Overflo PC If Overflo PC	= adv on w = 0; = adv w = 1;	dress (HERE dress (Jump dress (HERE))		Before Ir PC After Ins If Zo If Zo

BNZ Branch if Not Zero							
Synta	ax:	BNZ n					
Oper	ands:	-128 ≤ n ≤ 1	127				
Oper	ation:		if Zero bit is '0', (PC) + 2 + 2n \rightarrow PC				
Statu	s Affected:	None					
Enco	ding:	1110	0001 nn	nn nnnn			
Description:		If the Zero I will branch.	If the Zero bit is '0', then the program will branch.				
	The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a two-cycle instruction.						
Words: 1							
Cycle	es:	1(2)					
Q C If Ju	ycle Activity:						
	Q1	Q2	Q3	Q4			
	Decode	Read literal 'n'	Process Data	Write to PC			
	No operation	No operation	No operation	No operation			
lf No	o Jump:						
	Q1	Q2	Q3	Q4			
	Decode	Read literal 'n'	Process Data	No operation			
<u>Exan</u>	nple:	HERE	BNZ Jump)			
Before Instruction PC = address (HERE) After Instruction If Zero = 0:							
	If Zero PC PC PC	= ad = 1;	dress (Jump)				

BRA		Unconditio	onal Branch						
Synta	ax:	BRA n							
Oper	ands:	-1024 ≤ n ≤	1023						
Oper	ation:	(PC) + 2 + 2	$2n \rightarrow PC$						
Statu	s Affected:	None	None						
Enco	ding:	1101	0nnn nn	nn nnnn					
Description: Add the 2's complement number '2n' the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is a two-cycle instruction.									
Word	ls:	1	1						
Cycle	es:	2							
QC	ycle Activity:								
	Q1	Q2	Q3	Q4					
	Decode	Read literal 'n'	Process Data	Write to PC					
	No operation	No operation	No operation	No operation					
	n <u>ple:</u> Before Instruc PC After Instructio PC	= ad	BRA Jump dress (HERE dress (Jump)					

BSF	Bit Set f					
Syntax:	BSF f, b	{,a}				
Operands:	$0 \le f \le 255$ $0 \le b \le 7$ $a \in [0,1]$	$0 \le b \le 7$				
Operation:	$1 \rightarrow \text{f}$					
Status Affected:	None					
Encoding:	1000	bbba	ffff	ffff		
Description:	Bit 'b' in reg	Bit 'b' in register 'f' is set.				
	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.					
	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.					
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3		Q4		
Decode	Read register 'f'	Proces Data		Write gister 'f'		

Example:

Before Instruction FLAG_REG =

= 0Ah

FLAG_REG, 7, 1

After Instruction FLAG_REG = 8Ah

BSF

BTFSC	Bit Test File	, Skip if Clear		BTFSS		Bit Test File	, Skip if Set	
Syntax:	BTFSC f, b	{,a}		Syntax:		BTFSS f, b {	,a}	
Operands:	$0 \le f \le 255$ $0 \le b \le 7$ $a \in [0,1]$			Operand	ls:	0 ≤ f ≤ 255 0 ≤ b < 7 a ∈ [0,1]		
Operation:	skip if (f)	= 0		Operatio	n:	skip if (f)	= 1	
Status Affected:	None			Status A	ffected:	None		
Encoding:	1011	bbba ff	ff ffff	Encodin	g:	1010	bbba ffi	ff ffff
Description:	instruction is the next inst current instru and a NOP is	gister 'f' is '0', t skipped. If bit ruction fetched uction executio executed instruction.	'b' is '0', then during the in is discarded ead, making	Descript	ion:	instruction is the next instruction current instruction and a NOP is	gister 'f' is '1', t skipped. If bit uction fetched iction execution executed inste cle instruction.	'b' is '1', then during the n is discarded
		e Access Banł BSR is used to	k is selected. If a select the				e Access Bank BSR is used to	
	is enabled, t Indexed Lite whenever f ≤ Section 26.2 Bit-Oriented	d the extended his instruction ral Offset Addr 95 (5Fh). See 2.3 "Byte-Orie I Instructions et Mode" for d	essing mode ented and in Indexed			set is enable Indexed Liter whenever f ≤ Section 26.2 Bit-Oriented	d the extended d, this instruction ral Offset Addro 95 (5Fh). See 2.3 "Byte-Orient I Instructions of Mode" for de	on operates in essing mode nted and in Indexed
Words:	1			Words:		1		
Cycles:	•	cles if skip and 2-word instruc		Cycles:			vcles if skip and a 2-word instru	
Q Cycle Activity:				Q Cycle	e Activity:			
Q1	Q2	Q3	Q4		Q1	Q2	Q3	Q4
Decode	Read	Process	No		Decode	Read	Process	No
lf skip:	register 'f'	Data	operation	lf skip:		register 'f'	Data	operation
li skip. Q1	Q2	Q3	Q4	li skip.	Q1	Q2	Q3	Q4
No	No	No	No		No	No	No	No
operation	operation	operation	operation	C	peration	operation	operation	operation
If skip and followe	,			lf skip a		by 2-word inst		
Q1	Q2	Q3	Q4		Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation		No peration	No operation	No operation	No operation
No	No	No	No		No	No	No	No
operation	operation	operation	operation	c	peration	operation	operation	operation
Example:	HERE BI FALSE : TRUE :	IFSC FLAG	;, l, O	<u>Example</u>	<u>::</u>	HERE BI FALSE : TRUE :	FSS FLAG	, 1, 0
Before Instruc PC After Instructio If FLAG< PC	= add on :1> = 0;	ress (HERE)			ore Instruct PC er Instructio If FLAG< PC	= add n 1> = 0;	ress (HERE)	
If FLAG< PC		ress (False)		If FLAG< PC	1> = 1;	ress (TRUE)	

		Branon in e	Overflow	
BTG f, b {,a}	Syntax:	BOV n		
$0 \le f \le 255$	Operands:	-128 ≤ n ≤ 1	27	
0 ≤ b < 7 a ∈ [0,1]	Operation:		,	
$\overline{(f < b^{>})} \rightarrow f < b^{>}$	Status Affected:	None		
None	Encoding:	1110	0100 nn	nn nnnn
Encoding: 0111 bbba ffff ffff Description: Bit 'b' in data memory location 'f' is inverted			,	hen the
If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank. If 'a' is '0' and the extended instruction	n added to the PC. Sin added to the PC. Sin incremented to fetch instruction, the new PC + 2 + 2n. This in two-cycle instruction		e PC. Since the d to fetch the the new addro n. This instruc	e PC will have next ess will be
in Indexed Literal Offset Addressing	Words:	1		
mode whenever $f \le 95$ (5Fh). See	Cycles:	1(2)		
Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.	Q Cycle Activity: If Jump:			
1	Q1	Q2	Q3	Q4
1	Decode	Read literal 'n'	Process Data	Write to PC
	No	No	No	No
		operation	operation	operation
	•	00	00	Q4
				No
BTG PORTC, 4, 0	Decode	'n'	Data	operation
= 0111 0101 [75h] m:	PC After Instruction If Overflor PC	= ade on ow = 1; = ade)
	$\begin{array}{l} 0 \leq f \leq 255\\ 0 \leq b < 7\\ a \in [0,1]\\ (\overline{f < b >}) \rightarrow f < b >\\ \hline \\ \hline \\ None\\ \hline \hline \\ \hline \\ 0111 bbba ffff ffff\\ \hline \\ \\ \hline \\ Bit 'b' in data memory location 'f' is inverted.\\ \hline \\ If 'a' is '0', the Access Bank is selected.\\ \hline \\ If 'a' is '1', the BSR is used to select the GPR bank.\\ \hline \\ If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f \leq 95 (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.\\ 1\\ 1\\ \hline \\ Q2 \qquad Q3 \qquad Q4\\ \hline \\ \hline \\ Read \qquad Process \qquad Write \\ register 'f' \qquad Data \qquad register 'f'\\ \hline \\ BTG \qquad PORTC, \ 4, \ 0\\ \hline \\ tion: = 0111 \ 0101 \ [75h] \\ on: \end{array}$	$\begin{array}{cccccccccccccccccccccccccccccccccccc$	$\begin{array}{cccccccccccccccccccccccccccccccccccc$	$\begin{array}{cccccccccccccccccccccccccccccccccccc$

Syntax:BZnOperands: $-128 \le n \le 127$ Operation:if Zero bit is '1', (PC) + 2 + 2n \rightarrow PCStatus Affected:NoneEncoding: 1110 0000 nnnnDescription:If the Zero bit is '1', then the program will branch.The 2's complement number '2n' is added to the PC. Since the PC will ha incremented to fetch the next instruction, the new address will be PC + 2 + 2 n. This instruction is then two-cycle instruction.Words:1Cycles:1(2)Q Cycle Activity: If Jump:Q1Q1Q2Q3Q4NoNoNoNoNoNoNoQ1Q2Q3Q4	n ave					
Operation:if Zero bit is '1', (PC) + 2 + 2n \rightarrow PCStatus Affected:NoneEncoding:1110 0000 nnnn nnnDescription:If the Zero bit is '1', then the program will branch.The 2's complement number '2n' is added to the PC. Since the PC will have 	n ave					
$\begin{array}{c c} (PC) + 2 + 2n \rightarrow PC \\ \hline \\ \mbox{Status Affected:} None \\ \hline \\ \mbox{Encoding:} & \hline 1110 & 0000 & nnnn & nnn \\ \hline \\ \mbox{Encoding:} & \hline 1110 & 0000 & nnnn & nnn \\ \hline \\ \mbox{Description:} & \mbox{If the Zero bit is '1', then the program will branch.} \\ \hline \\ \mbox{The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then two-cycle instruction. \\ \hline \\ \mbox{Words:} & 1 \\ \hline \\ \mbox{Cycles:} & 1(2) \\ \hline \\ \mbox{Q Cycle Activity:} \\ \mbox{If Jump:} \\ \hline \\ \hline \\ \hline \\ \mbox{No} & No & No \\ \hline \\ \hline \\ \mbox{No} & No & No \\ \hline \\ \mbox{operation} & \mbox{operation} & \mbox{operation} \\ \hline \\ \mbox{If No Jump:} \\ \hline \end{array}$	n ave					
Encoding: 1110 0000 nnnn nnn Description: If the Zero bit is '1', then the program will branch. The 2's complement number '2n' is added to the PC. Since the PC will ha incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then two-cycle instruction. Words: 1 Cycles: 1(2) Q Cycle Activity: If Jump: Q1 Q2 Q3 Q4 Decode Read literal Process Write to 'n' No No No No No No No No If No Jump: Unite No Operation Operation	n ave					
Description: If the Zero bit is '1', then the program will branch. The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then two-cycle instruction. Words: 1 Cycles: 1(2) Q Cycle Activity: If Jump: Q1 Q2 Q3 Q4 Decode Read literal Process Write to 'n' No No No No No No No No If No Jump: Uncertain operation Operation Operation	n ave					
will branch. The 2's complement number '2n' is added to the PC. Since the PC will he incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then two-cycle instruction. Words: 1 Cycles: 1(2) Q Cycle Activity: If Jump: <u>Q1 Q2 Q3 Q4</u> <u>Decode Read literal Process Write to</u> <u>'n' Data PC</u> <u>No No No No No</u> operation operation operation If No Jump:	ave					
$\begin{array}{c c} \mbox{added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then two-cycle instruction. \\ \hline PC + 2 + 2n. This instruction is then two-cycle instruction. \\ \hline \hline Vords: 1 \\ \hline Cycles: 1(2) \\ Q Cycle Activity: \\ \mbox{If Jump:} \\ \hline \hline Q1 & Q2 & Q3 & Q4 \\ \hline \hline \hline Decode & Read literal & Process & Write to Construct on the ext of t$						
Cycles: 1(2) Q Cycle Activity: If Jump: Q1 Q2 Q3 Q4 Decode Read literal Process Write to 'n' Data PC No No No No operation operation operation If No Jump:						
Q Cycle Activity: If Jump: Q1 Q2 Q3 Q4 Decode Read literal Process Write to 'n' Data PC No No No No operation operation operation If No Jump:	Words: 1					
If Jump: Q1 Q2 Q3 Q4 Decode Read literal Process Write to 'n' Data PC No No No No operation operation operation If No Jump:	Cycles: 1(2)					
Decode Read literal (n') Process Data Write to PC No No No No operation operation operation operation If No Jump: Units Units Units						
'n' Data PC No No No No operation operation operation operation If No Jump: Vinter State Vinter State Vinter State						
No No No No operation operation operation operation						
operation operation operation operation						
If No Jump:	h					
	<u> </u>					
Decode Read literal Process No						
'n' Data operation	ı					
Example: HERE BZ Jump						
Before Instruction						
	PC = address (HERE)					
After Instruction If Zero = 1;						
PC = address (Jump)						
If Zero = 0; PC = address (HERE + 2)						

Syntax:	CALL k {,s	s}			
Operands:	0 ≤ k ≤ 104 s ∈ [0,1]				
Operation:	$(PC) + 4 \rightarrow k \rightarrow PC < 20$ if s = 1 (W) \rightarrow WS (STATUS) \rightarrow (BSR) \rightarrow B):1>; → STATU	JSS,		
Status Affected:	None				
Encoding: 1st word (k<7:0>) 2nd word(k<19:8>)	1110 1111	110s k ₁₉ kkk	k ₇ k kkł		kkkk kkkk
	(PC+ 4) is p If 's' = 1, th registers ar	e W, STA re also pu	ATUS ushed	and I	BSR
	respective STATUSS a update occ is loaded in two-cycle ir	and BSR urs. Ther ito PC<2	S. If 's n, the 2 0:1>.	s' = 0 20-bi), no it value
Words:	STATUSS a update occ	and BSR urs. Ther ito PC<2	S. If 's n, the 2 0:1>.	s' = 0 20-bi), no it value
Words: Cycles:	STATUSS a update occ is loaded in two-cycle ir	and BSR urs. Ther ito PC<2	S. If 's n, the 2 0:1>.	s' = 0 20-bi), no it value
	STATUSS a update occ is loaded in two-cycle in 2	and BSR urs. Ther ito PC<2	S. If 's n, the 2 0:1>.	s' = 0 20-bi), no it value
Cycles:	STATUSS a update occ is loaded in two-cycle in 2	and BSR urs. Ther ito PC<2	S. If 's n, the 2 0:1>. 0 n.	s' = 0 20-bi), no it value
Cycles: Q Cycle Activity:	STATUSS a update occ is loaded ir two-cycle ir 2 2	and BSR urs. Ther nto PC<2 nstructior	S. If 's n, the 2 0:1>. 1 n. 3 C to	S' = 0 20-bi CALI CALI), no it value _ is a Q4 ad litera <19:8>
Cycles: Q Cycle Activity: Q1	STATUSS a update occ is loaded ir two-cycle ir 2 2 Q2 Read literal	and BSR urs. Ther nto PC<2 nstruction Q3 Push P	S. If 's n, the 2 0:1>. n. 3 C to k	Rea 'k'<), no it value L is a
Cycles: Q Cycle Activity: Q1 Decode No	STATUSS a update occ is loaded in two-cycle in 2 2 Q2 Read literal 'k'<7:0>, No	and BSR urs. Ther nto PC<2 nstruction Q3 Push P stac No	S. If 's n, the 2 0:1>. 0:1>. 1 C to k	Rea 'k'<	Q4 ad litera <19:8> te to P No eratior
Cycles: Q Cycle Activity: Q1 Decode No operation	STATUSS a update occ is loaded in two-cycle in 2 2 2 Read literal 'k'<7:0>, No operation HERE	and BSR urs. Ther nto PC<2 nstruction Q3 Push P stac No operat	S. If 's n, the 2 0:1>. 0:1>. 1 C to k	Rea 'k'- Writ	Q4 ad litera <19:8> te to P No eratior
Cycles: Q Cycle Activity: Q1 Decode No operation	STATUSS a update occ is loaded in two-cycle in 2 2 2 Read literal 'k'<7:0>, No operation HERE	and BSR urs. Ther no PC<2 nstruction Q3 Push P stac No operat	S. If 's n, the 2 0:1> 0:1> n. C to k tion	Rea 'k'- Writ	Q4 ad litera <19:8> te to P No eratior

CLRF	Clear f					
Syntax:	CLRF f{,	CLRF f {,a}				
Operands:	0 ≤ f ≤ 255 a ∈ [0,1]					
Operation:	$\begin{array}{l} 000h \rightarrow f, \\ 1 \rightarrow Z \end{array}$					
Status Affected:	Z					
Encoding:	0110	101a	ffff	ffff		
Description:	Clears the register.	Clears the contents of the specified register.				
If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.						
	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.					
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3	1	Q4		
Decode	Read	Proce		Write		
	register 'f'	Data	a re	gister 'f'		
Example:	Example: CLRF FLAG_REG, 1					
Before Instruction FLAG_REG = 5Ah After Instruction FLAG_REG = 00h						

CLRWDT Clear Watchdog Timer						
Syntax:	CLRWDT					
Operands:	None					
Operation:						
Status Affected:	TO, PD					
Encoding:	0000	0000	0000	0100		
Description: CLRWDT instruction resets the Watchdog Timer. It also resets the post scaler of the WDT. Status bits, TO and PD, are set.				s <u>the</u> post-		
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3	5	Q4		
Decode	No operation	Proce Data		No peration		
Example: CLRWDT Before Instruction WDT Counter = ?						

Before Instruction		
WDT Counter	=	?
After Instruction		
WDT Counter	=	00h
WDT Postscaler	=	0
TO	=	1
PD	=	1

COMF	Compleme	ent f		CPF	SEQ	Compare f	with W, Skip	if f = W
Syntax:	COMF f	{,d {,a}}		Synt	ax:	CPFSEQ	f {,a}	
Operands:	$0 \leq f \leq 255$			Ope	rands:	0 ≤ f ≤ 255 a ∈ [0,1]		
	d ∈ [0,1] a ∈ [0,1]			One	ration:	(f) – (W),		
Operation:	$f \rightarrow dest$			ope		skip if (f) =	(W) comparison)	
Status Affected:	N, Z			Statu	us Affected:	None	, i i i i i i i i i i i i i i i i i i i	
Encoding:	0001	11da ff:	ff fff		oding:	0110	001a ffi	ff ffff
Description:	complemer stored in W	ts of register 'f nted. If 'd' is '0' /. If 'd' is '1', th < in register 'f'.	, the result is e result is		cription:	Compares f location 'f' t		f data memory of W by
	lf 'a' is '0', t	he Access Bai he BSR is use	nk is selected.			discarded a	en the fetched and a NOP is ex aking this a two	xecuted
	set is enabl in Indexed	nd the extende led, this instruct Literal Offset A	ction operates Addressing			,	he Access Bar he BSR is use	nk is selected. d to select the
	Section 26 Bit-Oriente	never f ≤ 95 (5) 5.2.3 "Byte-Or ed Instruction set Mode" for	iented and s in Indexed			set is enabl in Indexed	nd the extende ed, this instruc Literal Offset A lever f \leq 95 (51	ction operates Addressing
Words:	1						.2.3 "Byte-Or	
Cycles:	1					Bit-Oriente	ed Instruction set Mode" for	s in Indexed
Q Cycle Activity:	00	00	04	Wor	ds:	1		
Q1	Q2	Q3	Q4 Write to	Cycl		1(2)		
Decode	Read register 'f'	Process Data	destination	- , .		Note: 3 cy	cles if skip and 2-word instruc	
Example:	COME			QC	ycle Activity:			
Example:	COMF	REG, 0, 0			Q1	Q2	Q3	Q4
Before Instruct REG	ion = 13h				Decode	Read	Process	No
After Instructio				الأحا		register 'f'	Data	operation
REG	= 13h			lf sł	Q1	Q2	Q3	Q4
W	= ECh				No	No	No	No
					operation	operation	operation	operation
				lf sl	kip and followe	d by 2-word in	struction:	
					Q1	Q2	Q3	Q4
					No	No	No	No
					operation	operation	operation	operation
					No operation	No operation	No operation	No operation
				_	L •	• •	•	
				<u>Exar</u>	<u>nple:</u>	HERE NEQUAL EQUAL	CPFSEQ REG :	B, O
					Before Instruc			

=	HERE	
=	?	
=	?	
=	W;	
=	Address	(EQUAL)
≠	W;	
=	Address	(NEQUAL)
	= = =	= ? = ? = W; = Address ≠ W;

CPFS	SGT	Compare f	with W, Skip	if f > W			
Synta	ax:	CPFSGT	f {,a}				
Oper	ands:	$0 \le f \le 255$	0 ≤ f ≤ 255				
•		a ∈ [0,1]					
Oper	ation:	(f) - (W),					
•		skip if (f) >	(W)				
		(unsigned c	comparison)				
Statu	s Affected:	None					
Enco	ding:	0110	010a fff	f fff			
Desc	ription:	Compares t	he contents of	data memory			
		location 'f' t	o the contents an unsigned s	of the W by			
		If the conte	nts of 'f' are gro	eater than the			
		contents of	WREG, then t	he fetched			
			s discarded ar				
		executed in two-cycle ir	stead, making struction.	this a			
		lf 'a' is '0', t	he Access Bar	nk is selected.			
If 'a' is '1', the BSR is used to select GPR bank.							
If 'a' is '0' and the extended instructi							
set is enabled, this instruction operate							
in Indexed Literal Offset Addressing mode whenever f \leq 95 (5Fh). See							
			.2.3 "Byte-Ori				
			d Instruction				
		Literal Offs	set Mode" for	details.			
Word	ls:	1					
Cycle	es:	1(2)					
,		• •	ycles if skip ar	nd followed			
		by	a 2-word instru	uction.			
QC	ycle Activity:						
i	Q1	Q2	Q3	Q4			
	Decode	Read	Process	No			
lf sk	in:	register 'f'	Data	operation			
II SK	ip. Q1	Q2	Q3	Q4			
ĺ	No	No	No	No			
	operation	operation	operation	operation			
lf sk	ip and followe	d by 2-word in	struction:				
	Q1	Q2	Q3	Q4			
	No	No	No	No			
	operation	operation	operation	operation			
No No		No	No				
	operation	operation	operation	operation			
Exam	nole:	HERE	CPFSGT RE	G, 0			
Exam	1910.	NGREATER	:	0, 0			
		GREATER	:				
	Before Instruc	tion					
	PC		dress (HERE)			
	W After Instruction	= ?					
	After Instructic If REG						
	PC	= Ad	dress (GREAT	rer)			
	If REG	≤ W;					
	PC	= Ad	dress (NGREA	ATER)			

CPF	CPFSLT Compare f with W, Skip if f < W				
Synt	ax:	CPFSLT	f {,a}		
Ope	rands:	0 ≤ f ≤ 255 a ∈ [0,1]			
Ope	ration:	(f) – (W), skip if (f) < (unsigned o	(W) comparison)		
Statu	us Affected:	None			
Enco	oding:	0110	000a ff:	ff ffff	
Description: Compares the constraint of to the performing an up			to the contents	of W by	
		contents of instruction	nts of 'f' are le W, then the fe is discarded an istead, making instruction.	etched nd a NOP is	
			he Access Bai he BSR is use		
Word	ds:	1			
Cycl	es:		ycles if skip an a 2-word instru		
QC	cycle Activity:				
	Q1	Q2	Q3	Q4	
	Decode	Read register 'f'	Process Data	No operation	
lf sk		regiotor i	Dulu	oporation	
	Q1	Q2	Q3	Q4	
	No	No	No	No	
	operation	operation	operation	operation	
lf sk	kip and followe	d by 2-word in	struction:		
	Q1	Q2	Q3	Q4	
	No operation	No operation	No operation	No operation	
	No	No	No	No	
	operation	operation	operation	operation	
<u>Exar</u>	<u>nple:</u>	NLESS	CPFSLT REG, : :	1	
	Before Instruc				
	PC	= Ac	dress (HERE)	
	W After Instructio	= ?			
	After Instructio	on < Wi	:		
	PC	= Ac	dress (LESS)	
	If REG PC	≥ W = Ac	dress (NLES	S)	

DAW	Decimal A	djust W Regis	ster	DECF	Decrement	tf		
Syntax:	DAW			Syntax:	DECF f{,c	d {,a}}		
Operands:	None			Operands:	$0 \leq f \leq 255$			
Operation:	•	> 9] or [DC = 1 6 → W<3:0>;	a.,		$\begin{array}{l} d \in [0,1] \\ a \in [0,1] \end{array}$			
	else	,		Operation:	$(f) - 1 \rightarrow de$	est		
	(W<3:0>) –	→ W<3:0>		Status Affected:	C, DC, N, C	DV, Z		
	lf [W<7:4>	> 9] or [C = 1],	, then	Encoding:	0000	01da ff	ff ffff	
	$(W<7:4>) + 6 \rightarrow W<7:4>;$ C = 1; else $(W<7:4>) \rightarrow W<7:4>$		C = 1; else	result is sto	Decrement register 'f'. If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'.			
Status Affected:	(VV<7.42) - C	→ VV N /.4/					nk is selected.	
Encoding:	0000	0000 000	00 0111		lf 'a' is '1', t GPR bank.		ed to select the	
Description:	resulting fro variables (e and produc result.	ts the eight-bit om the earlier a each in packed es a correct pa	addition of two BCD format)		set is enabl in Indexed mode wher Section 26	led, this instru Literal Offset never f ≤ 95 (5 . 2.3 "Byte-O i	Fh). See	
Words:	1					set Mode" for		
Cycles:	1			Words:	1			
Q Cycle Activity:				Cycles:	1			
Q1	Q2	Q3	Q4	Q Cycle Activity:				
Decode	Read register W	Process Data	Write W	Q1	Q2	Q3	Q4	
Example 1:	DAW		·	Decode	Read register 'f'	Process Data	Write to destination	
Before Instruc				Evennley				
W C DC After Instructi W C DC	= A5h = 0 = 0 on = 05h = 1 = 0			Example: Before Instru CNT Z After Instruc CNT Z	uction = 01h = 0 tion = 00h	CNT, 1, C	1	
Example 2:				Z	= 1			
Before Instruc	ction							
W C DC After Instructi	= CEh = 0 = 0							
W	= 34h							
C DC	= 1 = 0							

DEC	DECFSZ Decrement f, Skip if 0					
Synta	ax:	DECFSZ f	⁺ {,d {,a}}			
Oper	ands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$				
Oper	ation:	(f) – $1 \rightarrow de$ skip if result				
Statu	is Affected:	None				
Enco	oding:	0010	11da fff	f ffff		
Desc	cription:	The contents of register 'f' are decremented. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in gister 'f'.				
If the result is '0', the next instruction which is already fetched is discarded and a NOP is executed instead, making it a two-cycle instruction.						
If 'a' is '0', the Access Bank is selected If 'a' is '1', the BSR is used to select the GPR bank.						
If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.						
Word	ds:	1				
Cycle	es:	•	cles if skip an 2-word instru			
QC	ycle Activity:					
	Q1	Q2	Q3	Q4		
	Decode	Read register 'f'	Process Data	Write to destination		
lf sk	ip:	<u> </u>	ļ	ļJ		
	Q1	Q2	Q3	Q4		
	No	No	No	No		
	operation	operation	operation	operation		
IT SK	Q1	d by 2-word in		04		
I	No	Q2 No	Q3 No	Q4 No		
	operation	operation	operation	operation		
	No	No	No	No		
operation operation operation operation						
Example: HERE DECFSZ CNT, 1, 1 GOTO LOOP						
CONTINUE Before Instruction						
	PC After Instruction	= Address				
	PC After Instructio CNT If CNT	= Address on = CNT - ^ = 0;	1			
	PC After Instructio CNT	= Address on = CNT – 1	1)		

DCF	SNZ	Decrement	Decrement f, Skip if Not 0						
Syntax:		DCFSNZ f {,d {,a}}							
Operands:		$\begin{array}{l} 0 \leq f \leq 255 \\ d \in [0,1] \\ a \in [0,1] \end{array}$	d ∈ [0,1]						
Operation:		$(f) - 1 \rightarrow dest,$ skip if result $\neq 0$							
Statu	is Affected:	None							
Enco	oding:	0100	0100 11da ffff ffff						
Description:		decremente placed in W	The contents of register 'f' are decremented. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f'.						
		instruction discarded a	If the result is not '0', the next instruction which is already fetched is discarded and a NOP is executed instead, making it a two-cycle instruction.						
		If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.							
		If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.							
Words: 1									
Cycle	es:								
QC	ycle Activity:								
	Q1	Q2	Q3	Q4					
	Decode	Read	Process	Write to					
lf old	, in .	register 'f'	Data	destination					
lf sk	up. Q1	Q2	Q3	Q4					
	No	No	No	No					
	operation	operation	operation	operation					
lf sk	ip and followe	d by 2-word in	struction:						
	Q1	Q2	Q3	Q4					
	No	No	No	No					
	operation	operation	operation	operation					
	No	No	No	No					
	operation	operation	operation	operation					
Example: HERE DCFSNZ TEMP, 1, 0 ZERO : NZERO :									
Before Instruction									
	TEMP After Instruction	= n	?						
	TEMP	=	= TEMP – 1,						
	If TEMP	=	= 0;						
	PC If TEMP	= ≠							
PC = Address (NZERO)									

GOTO		Unconditi	Unconditional Branch						
Synta	ax:	GOTO k	GOTO k						
Operands:		$0 \le k \le 104$	$0 \leq k \leq 1048575$						
Operation:		$k \rightarrow PC<2$	$k \rightarrow PC<20:1>$						
Status Affected:		None	None						
Encoding: 1st word (k<7:0>) 2nd word(k<19:8>)		1110 1111	1111 k ₁₉ kkk	k ₇ kk kkk		kkkk ₀ kkkk ₈			
Desc	ription:	anywhere range. The PC<20:1>.	GOTO allows an unconditional branch anywhere within entire 2-Mbyte memory range. The 20-bit value 'k' is loaded into PC<20:1>. GOTO is always a two-cycle instruction.						
Words:		2	2						
Cycles:		2	2						
Q Cycle Activity:									
	Q1	Q2 Q3			Q4				
	Decode	Read literal 'k'<7:0>,	No operat		'k'•	ad literal <19:8>, te to PC			
	No	No	No			No			
	operation	operation	operat	ion	ор	eration			
<u>Exam</u>	n <u>ple:</u> After Instructic PC =		RE THERE)						

INCF	Increment	Increment f							
Syntax:	INCF f{,	INCF f {,d {,a}}							
Operands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$	d ∈ [0,1]							
Operation:	(f) + 1 \rightarrow d	(f) + 1 \rightarrow dest							
Status Affected:	C, DC, N, OV, Z								
Encoding:	0010	10da ffff ff							
Description:	incremente placed in V	The contents of register 'f' are incremented. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f'.							
	lf 'a' is '1', f	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.							
	set is enab in Indexed mode when Section 26 Bit-Oriente	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.							
Words:	1								
Cycles:	1								
Q Cycle Activity:									
Q1	Q2	Q3			Q4				
Decode	Read	Proce			rite to				
	register 'f'	Data	1	des	tination				
Example:	INCF	CNT,	1, 0						
Before Instruc CNT Z DC After Instructio CNT Z C DC	= FFh = 0 = ? = ?								

INCF	SZ	Increment	f, Skip if	0		
Synta	ax:	INCFSZ f	{,d {,a}}			
Opera	ands:	$\begin{array}{l} 0 \leq f \leq 255 \\ d \in [0,1] \\ a \in [0,1] \end{array}$				
Opera	ation:	(f) + 1 \rightarrow de skip if result	-			
Statu	s Affected:	None				
Enco	ding:	0011	11da	ffff	ffff	
Desc	ription:	incremente placed in W	The contents of register 'f' are incremented. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f'.			
If the result is '0', the next instruction which is already fetched is discarded and a NOP is executed instead, making it a two-cycle instruction.					carded	
	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.					
		If 'a' is '0' a set is enabl in Indexed I mode when Section 26	ed, this in Literal Off lever $f \le 9$	struction set Addre 5 (5Fh).	operates essing See	
		Bit-Oriente Literal Offs				
Word	s:					
Cycle	es:	Literal Offs 1 1(2) Note: 3 c		" for deta kip and fo	ils. bllowed	
Cycle	es: ycle Activity:	Literal Offs 1 1(2) Note: 3 c by	set Mode cycles if sl a 2-word	" for deta kip and fo	ils. bllowed bn.	
Cycle	es:	Literal Offs 1 1(2) Note: 3 c	set Mode	i for deta kip and fo instructio	ils. bllowed	
Cycle	es: ycle Activity: Q1 Decode	Literal Offs 1 1(2) Note: 3 c by Q2 Read	et Mode cycles if sl a 2-word Q3 Proces	i for deta kip and fo instructio	ils. blowed bn. Q4 Vrite to	
Cycle Q Cy	es: ycle Activity: Q1 Decode ip: Q1	Literal Offs 1 1(2) Note: 3 c by Q2 Read	et Mode cycles if sl a 2-word Q3 Proces	i for deta kip and fo instructio	ils. blowed bn. Q4 Vrite to	
Cycle Q Cy	es: ycle Activity: Q1 Decode ip: Q1 No	Literal Offs 1 1(2) Note: 3 c by Q2 Read register 'f' Q2 No	et Mode cycles if sl a 2-word Q3 Proces Data Q3 No	ip and fo instructions v des	ils. ollowed on. Q4 Vrite to stination Q4 No	
Cycle Q Cy If ski	es: ycle Activity: Q1 Decode ip: Q1 No operation	Literal Offs 1 1(2) Note: 3 c by Q2 Read register 'f' Q2 No operation	et Mode cycles if sl a 2-word Q3 Proces Data Q3 No operatio	ip and fo instructions v des	ils. bllowed on. Q4 Vrite to stination Q4	
Cycle Q Cy If ski	es: ycle Activity: Q1 Decode ip: Q1 No operation ip and followe	Literal Offs 1 1(2) Note: 3 c by Q2 Read register 'f' Q2 No operation d by 2-word in:	expeles if sl a 2-word Q3 Proces Data Q3 No operations	ip and fo instructions v des	ils. ollowed on. Q4 Vrite to stination Q4 No peration	
Cycle Q Cy If ski	es: ycle Activity: Q1 Decode ip: Q1 No operation	Literal Offs 1 1(2) Note: 3 c by Q2 Read register 'f' Q2 No operation	et Mode cycles if sl a 2-word Q3 Proces Data Q3 No operatio	ip and fo instructions v des	ils. ollowed on. Q4 Vrite to stination Q4 No	
Cycle Q Cy If ski	es: ycle Activity: Q1 Decode ip: Q1 No operation ip and followe Q1	Literal Offs 1 1(2) Note: 3 c by Q2 Read register 'f' Q2 No operation d by 2-word in: Q2	et Mode cycles if sl a 2-word Q3 Proces Data Q3 No operatio struction: Q3	instructions v	ils. ollowed on. Q4 Write to stination Q4 No peration Q4	
Cycle Q Cy If ski	es: ycle Activity: Q1 Decode ip: Q1 No operation ip and followe Q1 No operation No operation No	Literal Offs 1 1(2) Note: 3 c by Q2 Read register 'f' Q2 No operation d by 2-word in: Q2 No operation No operation No	et Mode cycles if sl a 2-word Q3 Proces Data Q3 No operation: Q3 No operation: Q3 No operation: Q3	instructions v v v v v v v v v v v v v v v v v v v	ils. ollowed on. Q4 Vrite to stination Q4 No peration No No	
Cycle Q Cy If ski	es: ycle Activity: Q1 Decode ip: Q1 No operation ip and followe Q1 No operation	Literal Offs 1 1(2) Note: 3 c by Q2 Read register 'f' Q2 No operation d by 2-word in: Q2 No operation	et Mode cycles if sl a 2-word Q3 Proces Data Q3 No operation: Q3 No operation: Q3 No	instructions v v v v v v v v v v v v v v v v v v v	ils. ollowed on. Q4 Vrite to stination Q4 No peration Q4 No peration	
Cycle Q Cy If ski	es: ycle Activity: Q1 Decode ip: Q1 No operation ip and followe Q1 No operation No operation	Literal Offs 1 1(2) Note: 3 c by Q2 Read register 'f' Q2 No operation d by 2-word in: Q2 No operation No operation No operation HERE NZERO	et Mode cycles if sl a 2-word Q3 Proces Data Q3 No operation: Q3 No operation: Q3 No operation: Q3	instructions v v v v v v v v v v v v v v v v v v v	ils. ollowed on. Q4 Vrite to stination Q4 No peration No peration	
Cycle Q Cy If ski If ski Exam	es: ycle Activity: Q1 Decode ip: Q1 No operation ip and followe Q1 No operation No operation nple: Before Instruct PC	Literal Offs 1 1(2) Note: 3 c by Q2 Read register 'f' Q2 No operation d by 2-word in: Q2 No operation No operation HERE NZERO ZERO tion = Address	et Mode cycles if sl a 2-word Q3 Proces Data Q3 No operation: Q3 No operation: Q3 No operation: No operation: No operation: No	instructions v v v v v v v v v v v v v v v v v v v	ils. ollowed on. Q4 Vrite to stination Q4 No peration No peration	
Cycle Q Cy If ski If ski Exam	es: ycle Activity: Q1 Decode ip: Q1 No operation ip and followe Q1 No operation No operation nple: Before Instruction PC After Instruction	Literal Offs 1 1(2) Note: 3 c by Q2 Read register 'f' Q2 No operation d by 2-word in: Q2 No operation Mo operation HERE NZERO ZERO tion = Address n = CNT + 4	et Mode cycles if sl a 2-word Q3 Proces Data Q3 No operation: Q3 No operation: No No No No No No No No No No	instructions v v v v v v v v v v v v v v v v v v v	ils. ollowed on. Q4 Vrite to stination Q4 No peration No peration	
Cycle Q Cy If ski If ski	es: ycle Activity: Q1 Decode ip: Q1 No operation ip and followe Q1 No operation No operation nple: Before Instruct PC After Instruction	Literal Offs 1 1(2) Note: 3 c by Q2 Read register 'f' Q2 No operation d by 2-word in: Q2 No operation No operation No operation HERE IZERO ZERO tion = Address on = CNT + 7	et Mode cycles if sl a 2-word Q3 Proces Data Q3 No operation: Q3 No operation: No No No No No No No No No No	instructions v v v v v v v v v v v v v v v v v v v	ils. ollowed on. Q4 Vrite to stination Q4 No peration No peration	

INFS	SNZ	Increment	f, Skip if Not	0		
Synt	ax:	INFSNZ f	{,d {,a}}			
Oper	rands:	$0 \leq f \leq 255$				
		d ∈ [0,1]				
		a ∈ [0,1]				
Oper	ration:	(f) + 1 \rightarrow de skip if result				
Statu	is Affected:	None				
Enco	oding:	0100	10da ffi	Ef ffff		
Desc	cription:	The content	ts of register 'f	' are		
		placed in W	d. If 'd' is '0', tł /. If 'd' is '1', th ‹ in register 'f'.	e result is		
		If the result is not '0', the next instruction which is already fetched is discarded and a NOP is executed instead, making it a two-cycle instruction.				
			he Access Bar he BSR is use			
	GPR bank. If 'a' is '0' and the extended instruction set is enabled, this instruction operate in Indexed Literal Offset Addressing mode whenever $f \le 95$ (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed					
Word	1e.	1	set Mode" for	actano.		
Cycle		1(2)				
Cyck		Note: 3 cy	cles if skip an 2-word instru			
QC	cycle Activity:	·				
	Q1	Q2	Q3	Q4		
	Decode	Read	Process	Write to		
		register 'f'	Data	destination		
lf sk	kip:					
	Q1	Q2	Q3	Q4		
	No	No	No	No		
lf ol	operation	operation	operation	operation		
11 56	kip and followe Q1	Q2	Q3	Q4		
	No	No	No	No		
	operation	operation	operation	operation		
	No	No	No	No		
	operation	operation	operation	operation		
<u>Exar</u>	<u>nple:</u>	HERE I ZERO NZERO	INFSNZ REG	s, 1, 0		
	Before Instruction					
	PC		(HERE)			
	After Instructio		1			
	REG If REG	= REG + ⁷ ≠ 0;	I			
	PC If REG	 Address 	(NZERO)			
	PC	= 0; = Address	(ZERO)			

IORLW	Inclusive OR Literal with W				
Syntax:	IORLW k				
Operands:	$0 \le k \le 255$	5			
Operation:	(W) .OR. k	$\rightarrow W$			
Status Affected:	N, Z				
Encoding:	0000 1001 kkkk kkkk				
Description:	The conter eight-bit lite in W.				
Words:	1				
Cycles:	1				
Q Cycle Activity:					
Q1	Q2	Q3		Q4	
Decode	Read	Proce		Vrite to	
	literal 'k'	Data	a	W	
Example:	IORLW	35h			
Before Instruc W After Instructio	= 9Ah				

ter Instruc	tion	
W	=	BFh

IORWF	Inclusive C	R W with f			
Syntax:	IORWF f	{,d {,a}}			
Operands:	$\begin{array}{l} 0 \leq f \leq 255 \\ d \in [0,1] \\ a \in [0,1] \end{array}$				
Operation:	(W) .OR. (f) \rightarrow dest				
Status Affected:	N, Z				
Encoding:	0001	00da ff	ff ffff		
Description:	Inclusive OR W with register 'f'. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f'.				
	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.				
	set is enabl in Indexed I mode when Section 26 Bit-Oriente	nd the extend ed, this instru- Literal Offset λ ever f \leq 95 (5 .2.3 "Byte-Or d Instruction set Mode" for	ction operates Addressing Fh). See iented and s in Indexed		
Words:	1				
Cycles:	1				
Q Cycle Activity:					
Q1	Q2	Q3	Q4		
Decode	Read register 'f'	Process Data	Write to destination		
Example:		Data			

impic.	τu	ICMT.
Before Instruc	tion	
RESULT	=	13h
W	=	91h
After Instruction	n	
RESULT	=	13h
W	=	93h

LFSF	र	Load FSR					
Synta	ax:	LFSR f, k	LFSR f, k				
Operands:		$\begin{array}{l} 0 \leq f \leq 2 \\ 0 \leq k \leq 408 \end{array}$	$\begin{array}{l} 0 \leq f \leq 2 \\ 0 \leq k \leq 4095 \end{array}$				
Oper	ation:	$k\toFSRf$					
Statu	s Affected:	None					
Enco	ding:	1110 1111	1110 0000	00ff k ₇ kkk	k ₁₁ kkk kkkk		
Desc	ription:	The 12-bit file select r					
Words:		2					
Cycle	es:	2	2				
QC	ycle Activity:						
	Q1	Q2	Q3		Q4		
	Decode	Read literal 'k' MSB	Proce Data		Write literal 'k' MSB to FSRfH		
	Decode	Read literal 'k' LSB	Process Data		/rite literal ' to FSRfL		
Example: LFSR 2, 3ABh							
After Instruction FSR2H = 03h FSR2L = ABh							

MOVF	Move f					
Syntax:	MOVF f{	,d {,a}}				
Operands:	$\begin{array}{l} 0 \leq f \leq 255 \\ d \in [0,1] \\ a \in [0,1] \end{array}$					
Operation:	$f \to \text{dest}$					
Status Affected:	N, Z					
Encoding:	0101	00da	ffff	ffff		
Description:	The contents of register 'f' are moved to a destination dependent upon the status of 'd'. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f'. Location, 'f', can be anywhere in the 256-byte bank.					
	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.					
	If 'a' is '0' a set is enab in Indexed mode wher Section 26 Bit-Oriente Literal Offe	led, this i Literal O never f ≤ 5.2.3 "By ed Instru	nstruction ffset Add 95 (5Fh). te-Orient ctions ir	n operates ressing See ted and indexed		
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3	3	Q4		
Decode	Read register 'f'	Proce Data		Write W		
Example: Before Instruc		EG, 0,	0			
REG W After Instructio	= 22 = FF					
REG W	= 22 = 22					

MOVFF	Move f to f	: 			
Syntax:	MOVFF f _s	,f _d			
Operands:	$\begin{array}{l} 0 \leq f_s \leq 409 \\ 0 \leq f_d \leq 409 \end{array}$				
Operation:	$(f_s) \to f_d$				
Status Affected:	None				
Encoding: 1st word (source) 2nd word (destin.)	1100 ffff ffff ffff _s 1111 ffff ffff ffff _d				
Description:	The contents of source register ' f_s ' are moved to destination register ' f_d '. Location of source ' f_s ' can be anywhere in the 4096-byte data space (000h to FFFh) and location of destination ' f_d ' can also be anywhere from 000h to FFFh.				
	Either source or destination can be W (a useful special situation).				
	MOVFF is particularly useful for transferring a data memory location to a peripheral register (such as the transmit buffer or an I/O port).				
	The MOVFF PCL, TOSU destination	J, TOSH			
Words:	2				
Cycles:	2				
Q Cycle Activity:					
Q1	Q2	Q3		Q4	
Decode	Read register 'f' (src)	Proce Data		No peration	
Decode	No operation	No operat	ion re	Write gister 'f'	

Before Instruction		
REG1	=	33h
REG2	=	11h
After Instruction		
REG1	=	33h
REG2	=	33h

MOVLB	Move Liter	Move Literal to Low Nibble in BSR				
Syntax:	MOVLW k	(
Operands:	$0 \le k \le 255$					
Operation:	$k \to BSR$					
Status Affected:	None					
Encoding:	0000	0001	kkk	:k	kkkk	
Description:	Bank Select of BSR<7:4	The eight-bit literal 'k' is loaded into the Bank Select Register (BSR). The value of BSR<7:4> always remains '0' regardless of the value of k_7 : k_4 .				
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3	5		Q4	
Decode	Read literal 'k'	Proce Data			te literal to BSR	
Example:	MOVLB	5				
Before Instruc BSR Reg	gister = 02	'h				

After Instruction BSR Register = 05h

Move W to f

MOVWF

MOVLW Move Literal to W						
Syntax:	MOVLW I	<				
Operands:	$0 \le k \le 255$	5				
Operation:	$k \to W$					
Status Affected:	None					
Encoding:	0000	1110	kkkk	kkkk		
Description:	The eight-t	The eight-bit literal 'k' is loaded into W.				
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3		Q4		
Decode	Read literal 'k'	Proces Data		/rite to W		
Example:	MOVLW	5Ah				

After Instruction W

=

5Ah

Synta	ıx.	MOVWF	f {,a}					
		$0 \le f \le 255$						
Opera	anus.	0 ≤ 1 ≤ 255 a ∈ [0,1]						
Opera	ation:	$(W) \rightarrow f$						
•	s Affected:	None						
Enco	ding:	0110	111a	ffff	ffff			
Desc	ription:	Location 'f'	Move data from W to register 'f'. Location 'f' can be anywhere in the 256-byte bank.					
		If 'a' is '0', ' If 'a' is '1', ' GPR bank	the BSR is					
		set is enab in Indexed mode whe Section 26 Bit-Orient	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.					
Word	s:	1	1					
Cycle	s:	1						
QC	cle Activity:							
-	Q1	Q2	Q3		Q4			
	Decode	Read	Proces	ss	Write			
		register 'f'	Data	re	gister 'f'			
Exam	<u>iple:</u>	MOVWF	reg, 0					
I	Before Instruc	tion						
	W	= 4Fh						
	REG	= FFh						

After Instruction

W REG

4Fh 4Fh

= =

MULLW	Multiply Li	teral with W				
Syntax:	MULLW	MULLW k				
Operands:	$0 \le k \le 255$					
Operation:	(W) x k \rightarrow	PRODH:PROI	DL			
Status Affected:	None					
Encoding:	0000	1101 kki	kk kkkk			
Description:	out betwee 8-bit literal placed in th	An unsigned multiplication is carried out between the contents of W and the 8-bit literal 'k'. The 16-bit result is placed in the PRODH:PRODL register pair. PRODH contains the high byte.				
	W is uncha	nged.				
	None of the	e Status flags	are affected.			
	possible in	Note that neither Overflow nor Carry is possible in this operation. A Zero result is possible but not detected.				
Words:	1	1				
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3	Q4			
Decode	Read literal 'k'	Process Data	Write registers PRODH: PRODL			
Example:	MULLW	0C4h				
Before Instruc	tion					
W PRODH PRODL	= E2 = ? = ?	'n				
After Instructio W PRODH PRODL	on = E2 = AE = 08	Dh				

Syntax:	MULWF	f {,a}			
Operands:	$0 \le f \le 255$				
a	a ∈ [0,1]				
Operation:	(W) x (f) \rightarrow PRODH:PRODL				
Status Affected:	None		1	- 1	
Encoding:	0000	001a	ffff		
Description:	An unsigned multiplication is carried out between the contents of W and the register file location 'f'. The 16-bit result is stored in the PRODH:PRODL register pair. PRODH contains the high byte. Both W and 'f' are unchanged.				
	None of the Status flags are affected.				
	Note that no possible in possible bu	this opera	tion. A 2		
	If 'a' is '0', t 'a' is '1', the GPR bank.				
	If 'a' is '0' ar is enabled, Indexed Litr whenever f Section 26 Bit-Oriente Literal Offs	this instru eral Offse ≤ 95 (5Fh .2.3 "Byte ed Instruc	t Addres Addres See -Orien	berates in ssing mode ted and h Indexed	
Words:	1				
Cycles:	1				
Q Cycle Activity:					
Q1	Q2	Q	3	Q4	
Decode	Read register 'f'	Proc Da		Write registers PRODH: PRODL	
Example:	MULWF	REG,	1		
Before Instru	ction				
W REG PRODH	=	C4h B5h ? ?			

After Instruction

W REG PRODH PRODL C4h B5h 8Ah 94h

= = =

NEGF	Negate f					
Syntax:	NEGF f {,a}					
Operands:	0 ≤ f ≤ 255 a ∈ [0,1]					
Operation:	$(\overline{f}) + 1 \rightarrow f$					
Status Affected:	N, OV, C, DC, Z					
Encoding:	0110 110a ffff ffff					
Description: Location 'f' is negated using two's complement. The result is placed in data memory location 'f'.						
	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.					
	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \le 95$ (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.					
Words:	1					
Cycles:	1					
Q Cycle Activity:						
04	00 00 01					

Q1	Q2	Q3	Q4
Decode	Read	Process	Write
	register 'f'	Data	register 'f'

Example:	NEGF	REG,	1
----------	------	------	---

Before Instru	ction			
REG	=	0011	1010	[3Ah]
After Instructi	on			
REG	=	1100	0110	[C6h]

NOP		No Operation					
Synta	ax:	NOP					
Oper	ands:	None					
Oper	ation:	No operati	on				
Statu	s Affected:	None	None				
Encoding:		0000	0000	000	0	0000	
		1111	xxxx	XXX	x	XXXX	
Desc	ription:	No operation.					
Word	ls:	1					
Cycle	es:	1					
QC	ycle Activity:						
	Q1 Q2		Q3		Q4		
	Decode	No	No	No		No	
		operation	operation o			eration	

Example:

None.

POP	POP Pop Top of Return Stack						
Synta	x:	POP					
Opera	ands:	None					
Opera	ation:	$(TOS) \rightarrow b$	it bucke	et			
Status	Affected:	None					
Encoc	ling:	0000	0000	000	00	0110	
Descr		The TOS v stack and i then becon was pushe This instruc the user to stack to inc	s discai nes the d onto t ction is properl	rded. Th previou he retui provide y mana	ne T(is va rn sta d to ge th	DS value lue that ack. enable ne return	
Words	S:	1					
Cycles	s:	1					
Q Cy	cle Activity:						
_	Q1	Q2	0	Q3		Q4	
	Decode	No operation	POP val		ор	No eration	
<u>Exam</u>	<u>ple:</u>	POP GOTO	NEW				
Before Instruction TOS Stack (1 level down)		= =	0031A 014332				
A	After Instructic TOS PC	'n	= =	014332 NEW	2h		

PUS	н	Push Top o	of Ret	urn Stac	:k		
Synta	ax:	PUSH	PUSH				
Oper	ands:	None					
Oper	ation:	$(PC + 2) \rightarrow$	TOS				
Statu	s Affected:	None					
Enco	oding:	0000	0000	000	0	0101	
Desc	ription:	The PC + 2 the return s value is pus This instruc software sta then pushin	tack. ⁻ shed d tion al ack by	The prev own on lows imp modifyir	ious the s blem ng T	TOS stack. enting a OS and	
Word	ls:	1	1				
Cycle	es:	1	1				
QC	ycle Activity:						
	Q1	Q2	Q3			Q4	
	Decode	PUSH PC + 2 onto return stack		No ration	ор	No eration	
Exan		PUSH					
Before Instruction TOS PC			= =	345Ah 0124h			
	After Instructio PC TOS Stack (1	on level down)	= = =	0126h 0126h 345Ah			

RCALL	Relative Ca	Relative Call			
Syntax:	RCALL n				
Operands:	-1024 ≤ n ≤	1023			
Operation:	· · ·	$\begin{array}{l} (PC) + 2 \rightarrow TOS, \\ (PC) + 2 + 2n \rightarrow PC \end{array}$			
Status Affected:	None				
Encoding:	1101	1nnn	nnn	n	nnnn
Description: Words:	from the cu address (PC stack. Then number '2n' have incren instruction, PC + 2 + 2r	Subroutine call with a jump up to 1K from the current location. First, return address (PC + 2) is pushed onto the stack. Then, add the 2's complement number '2n' to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is a two-cycle instruction.			
Cycles:	2				
Q Cycle Activity:	-				
Q1	Q2	Q3	5		Q4
Decode	Read literal 'n' PUSH PC to stack	Proce Data		Writ	te to PC
No	No	No			No
operation	operation	operat	ion	ор	eration

Example: HERE RCALL Jump

Before Instruction

PC = Address (HERE) After Instruction PC = Address (Jump) TOS = Address (HERE + 2)

RESET Reset							
Synta	ax:	RESET					
Oper	ands:	None					
Operation:			Reset all registers and flags that are affected by a MCLR Reset.				
Statu	is Affected:	All					
Enco	oding:	0000	0000	1111	1111		
Desc	cription:		This instruction provides a way to execute a MCLR Reset in software.				
Word	ls:	1	1				
Cycle	es:	1					
QC	ycle Activity:						
	Q1	Q2	Q3		Q4		
	Decode	Start	No		No		
		reset	operat	ion op	peration		
		16361	operat				

Example:

After Instruction

Registers =	Reset Value
Flags* =	Reset Value

RESET

RET	FIE	Return from	m Interrupt		RET	LW	Return Lite	eral to W		
Synta	ax:	RETFIE {	\$}		Synt	ax:	RETLW k			
Oper	ands:	$s \in [0,1]$			Oper	rands:	$0 \le k \le 255$			
Oper	ation:	if s = 1,	IEH or PEIE/C	BIEL;	Oper	ration:	$k \rightarrow W$, (TOS) $\rightarrow P$ PCLATU, P	C, PCLATH are u	nchanged	
		$(WS) \rightarrow W,$			Statu	us Affected:	None			
		(STATUSS) (BSRS) \rightarrow	\rightarrow STATUS, BSR.		Enco	oding:	0000	1100 kk	kk kkkk	
		. ,	CLATH are ur	nchanged	Desc	cription:	W is loaded	d with the eigh	nt-bit literal 'k'.	
Statu	is Affected:	GIE/GIEH,	PEIE/GIEL.						baded from the	
	oding:	0000	0000 00					top of the stack (the return address). The high address latch (PCLATH)		
Desc	cription:		n interrupt. Sta Stack (TOS) is				remains un	changed.		
			errupts are en		Word		1			
		•	er the high or		Cycl		2			
		•	the shadow re	t. If 's' = 1, the	QC	cycle Activity:				
			and BSRS are	•		Q1	Q2	Q3	Q4	
			ponding regist			Decode	Read literal 'k'	Process Data	POP PC from stack,	
			id BSR. If 's' = gisters occurs.	0, no update			interear it	Data	write to W	
Word	le:	1				No	No	No	No	
		2				operation	operation	operation	operation	
Cycle		2			_					
QC	ycle Activity: Q1	Q2	Q3	Q4	Exar	<u>mple:</u>				
	Decode	No operation	No operation	POP PC from stack Set GIEH or GIEL		CALL TABLE	; W conta: ; offset v ; W now ha ; table va	value as		
	No	No	No	No	TABI					
	operation	operation	operation	operation		ADDWF PCL	; W = offs	set		
<u>Exan</u>	nple:	RETFIE	1			RETLW k0 RETLW k1	; Begin ta ;	able		
	After Interrupt					:				
	PC W BSR		= TOS = WS = BSRS			RETLW kn	; End of t	table		
	STATUS GIE/GIE	H, PEIE/GIEL	= STATU = 1	JSS		Before Instruc W	ction = 07h			
						After Instruction	0/11			
						W	 value of 	f kn		

RET	URN	Return from Subroutine			
Synta	ax:	RETURN	{s}		
Oper	ands:	$s \in [0,1]$			
Oper	ation:	$(TOS) \rightarrow PC;$ if s = 1, $(WS) \rightarrow W,$ $(STATUSS) \rightarrow STATUS,$ $(BSRS) \rightarrow BSR,$ PCLATU, PCLATH are unchanged			
Statu	s Affected:	None			
Enco	ding:	0000	0000	0001	001s
	ription:	Return from subroutine. The stack is popped and the top of the stack (TOS) is loaded into the program counter. If 's'= 1, the contents of the shadow registers WS, STATUSS and BSRS are loaded into their corresponding registers W, STATUS and BSR. If 's' = 0, no update of these registers occurs.			
Word	ls:	1			
Cycle	es:	2			
QC	ycle Activity:				
	Q1	Q2	Q	3	Q4
	Decode	No operation	Proce Dat		POP PC
	No	No	No)	No
	operation	operation	opera	tion o	peration
<u>Exan</u>	nple:	RETURN			

Example:	RE

After Instruction: PC = TOS

RLCF	Rotate Left	f through Ca	irry	
Syntax:	RLCF f {,	d {,a}}		
Operands:	$0 \le f \le 255$ d $\in [0,1]$ a $\in [0,1]$			
Operation:	$(f < n >) \rightarrow des$ $(f < 7 >) \rightarrow C,$ $(C) \rightarrow dest < 0$			
Status Affected:	C, N, Z			
Encoding:	0011	01da ff	ff ffff	
Description:	The contents of register 'f' are rotated one bit to the left through the Carry flag. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is stored back in register 'f'.			
			nk is selected. Id to select the	
	set is enable in Indexed Li mode whene Section 26.2 Bit-Oriented	d, this instruction iteral Offset A ever f ≤ 95 (5 2.3 " Byte-Or	⁻ h). See iented and s in Indexed	
	C ·	 register 	er f 🔫	
Words:	1			
o 1				
Cycles:	1			
Cycles: Q Cycle Activity:	1			
Cycles: Q Cycle Activity: Q1	1 Q2	Q3	Q4	
Q Cycle Activity:		Q3 Process Data	Q4 Write to destination	
Q Cycle Activity: Q1 Decode	Q2 Read	Process Data	Write to destination	
Q Cycle Activity: Q1 Decode Example:	Q2 Read register 'f'	Process	Write to destination	
Q Cycle Activity: Q1 Decode Example: Before Instruct REG C	Q2 Read register 'f' RLCF ction = 1110 (= 0	Process Data	Write to destination	
Q Cycle Activity: Q1 Decode Example: Before Instruct REG	Q2 Read register 'f' RLCF ction = 1110 (= 0 on = 1110 (Process Data REG, 0	Write to destination	

RLNCF	Rotate Left f (No Carry)	RRCF	Rotate Rig	ht f through	Carry
Syntax:	RLNCF f {,d {,a}}		Syntax:	RRCF f{,	d {,a}}	
Operands:	$\begin{array}{l} 0 \leq f \leq 255 \\ d \in [0,1] \\ a \in [0,1] \end{array}$		Operands:	$\begin{array}{l} 0 \leq f \leq 255 \\ d \in [0,1] \\ a \in [0,1] \end{array}$		
Operation:	$(f) \rightarrow dest,$ $(f<7>) \rightarrow dest<0>$		Operation:	$(f < n >) \rightarrow de$ $(f < 0 >) \rightarrow C$ $(C) \rightarrow dest$,	
Status Affected:	N, Z		Status Affected:	C, N, Z		
Encoding:		ff ffff	Encoding:	0011	00da ff	ff ffff
Description:	The contents of register one bit to the left. If 'd' is is placed in W. If 'd' is '2 stored back in register ' If 'a' is '0', the Access Ba	s '0', the result .', the result is ".	Description:	The conten one bit to th flag. If 'd' is If 'd' is '1', t	ts of register ' ne right throug	f' are rotated h the Carry is placed in W.
	If 'a' is '1', the BSR is us	ed to select the		register 'f'.		
	GPR bank. If 'a' is '0' and the exten set is enabled, this instri			,		nk is selected. ed to select the
	in Indexed Literal Offset mode whenever f ≤ 95 (Section 26.2.3 "Byte-O Bit-Oriented Instructio Literal Offset Mode" for register	5Fh). See priented and ins in Indexed or details.		set is enabl in Indexed mode wher Section 26 Bit-Oriente	led, this instru Literal Offset never f ≤ 95 (5 . 2.3 "Byte-O	Fh). See riented and ns in Indexed
Words:	1			− C	→ registe	er f 🗕 🍝
Cycles:	1					
Q Cycle Activity:			Words:	1		
Q1	Q2 Q3	Q4	Cycles:	1		
Decode	Read Process	Write to	Q Cycle Activity:			_
	register 'f' Data	destination	Q1	Q2	Q3	Q4
English			Decode	Read register 'f'	Process Data	Write to destination
Example:	RLNCF REG, 1,	0				
Before Instruc REG	= 1010 1011		Example:	RRCF	REG, 0,	0
After Instruction REG	on = 0101 0111		Before Instruc REG C After Instructi	= 1110 0 = 0)110	
			REG W C	on = 1110 (= 0111 (= 0		

RRN	ICF	Rotate Ri	ght f (No	Carry)
Synt	ax:	RRNCF	f {,d {,a}}		
Ope	rands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$	5		
Ope	ration:	$(f < n >) \rightarrow (f < 0 >) \rightarrow (f <$		1>,	
Statu	us Affected:	N, Z			
Enco	oding:	0100	00da	fff	f ffff
Deso	cription:	one bit to	the right. in W. If 'd'	lf 'd' is is '1',	are rotated '0', the result the result is
			overriding the bank	the BS	ik will be SR value. If 'a' e selected as
		set is ena in Indexed mode whe Section 2	bled, this d Literal O enever f ≤ c6.2.3 "By ted Instru	instruc ffset A 95 (5F r te-Ori e uctions	ented and s in Indexed
		Г	► re	egister	f
Wor	ds:	1			
Cycl		1			
,	cycle Activity:				
	Q1	Q2	Q	3	Q4
	Decode	Read register 'f'	Proce Dat		Write to destination
<u>Exar</u>	nple 1:	RRNCF	REG, 1	, 0	
	Before Instruc REG After Instructio	= 1101	0111		
	REG		1011		
Exar	nple 2:	RRNCF	REG, 0	, 0	
	Before Instruc W REG After Instructio	= ? = 1101	0111		
	w REG	= 1110 = 1101			

OFTE	0-44			
SETF	Set f			
Syntax:	SETF f {,a}			
Operands:	0 ≤ f ≤ 255 a ∈ [0,1]			
Operation:	$FFh\tof$			
Status Affected:	None			
Encoding:	0110 100a ffff ffff			
Description:	The contents of the specified register are set to FFh.			
	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.			
	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.			
Words:	1			
Cycles:	1			
Q Cycle Activity:				
Q1	Q2 Q3 Q4			
Decode	Read Process Write register 'f' Data register 'f'			
Example:	SETF REG,1			
Before Instruc REG After Instructio REG	= 5Ah			

SLE	EP	Enter Slee	ep Mode		SUB	FWB	Subtract f fr	om W with Bo	orrow
Synta	ax:	SLEEP			Synta	ax:	SUBFWB f	{,d {,a}}	
Oper	ands:	None			Oper	ands:	$0 \leq f \leq 255$		
Oper	ation:	$00h \rightarrow WE$	DT,				d ∈ [0,1]		
			postscaler,		0	-ti	a ∈ [0,1]		
		$1 \rightarrow \frac{\text{TO}}{\text{PD}},$ $0 \rightarrow \frac{\text{PD}}{\text{PD}}$			•	ation: s Affected:	$(W) - (f) - (\overline{C})$		
Statu	s Affected:	TO, PD					N, OV, C, DC		c
Enco	dina:	0000	0000 000	0 0011		ding:		01da fff	
Desc	ription:	0000 0000 0011 The Power-Down status bit (PD) is cleared. The Time-out status bit (TO) is set. The Watchdog Timer and its postscaler are cleared.		Desc	Description:	Subtract register 'f' and Carry flag (borrow) from W (2's complement method). If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored in register 'f'.			
Word	le.	•	ssor is put into scillator stoppe	•				e Access Bank BSR is used to	is selected. If select the
Cycle		1						d the extended	
	ycle Activity:							a, this instructional Offset Addr	on operates in essing mode
	Q1	Q2	Q3	Q4			whenever f \leq	95 (5Fh). See	;
	Decode	No operation	Process Data	Go to Sleep			Bit-Oriented	.3 "Byte-Orie Instructions t Mode" for d	in Indexed
Exan	nnle:	SLEEP			Word	ls:	1		
	Before Instruc				Cycle	es:	1		
	TO =	?			QC	ycle Activity:			
	PD =	?				Q1	Q2	Q3	Q4
	After Instructio TO = PD =	on 1† 0				Decode	Read register 'f'	Process Data	Write to destination
	10	0			Exan	nple 1:	SUBFWB	REG, 1, 0	
† If	WDT causes	wake-up, this b	oit is cleared.			Before Instruc REG	tion = 3		
						W	= 2 = 1		
						After Instruction	•		
						REG W	= FF = 2		
						С	= 0		
						Ž N	= 0 = 1 ;r	esult is negati	ve
					Exan	nple 2:	SUBFWB		
						Before Instruc			
						REG W	= 2 = 5		
						C After Instructio	= 1		
						REG	= 2		
						W C	= 3 = 1		
						Ž N	= 0	esult is positiv	0
					Evan	IN Anle 3:	= U ; f		

; result is zero

REG, 1, 0

Before Instruction REG W C

After Instruction

REG W C Z N

SUBFWB

1 2 0 = =

Example 3:

SUBLW	5	Subtrac	t ۱:	N from L	itera	I	
Syntax:	S	SUBLW	k	(
Operands:	C	$0 \le k \le 2$	255	5			
Operation:	k	x – (W)	\rightarrow	W			
Status Affected:	١	N, OV, 0	C, I	DC, Z			
Encoding:	Γ	0000		1000	kkk	ck	kkkk
Description:				acted from			
Words:	1						
Cycles:	1						
Q Cycle Activity:							
Q1		Q2		Q3		-	Q4
Decode		Read		Proces		V	Vrite to
	lite	eral 'k'		Data	l		W
Example 1:	S	SUBLW	С	2h			
Before Instruc W C After Instructio	=	01h ?					
W C Z N	= = = =	01h 1 0 0	;।	result is p	oositiv	/e	
Example 2:	S	SUBLW	C	2h			
Before Instruc	tion						
W C	=	02h ?					
After Instructio W C Z N	on = = = =	00h 1 1 0	;	result is z	zero		
Example 3:	S	SUBLW	C	2h			
Before Instruct W C After Instructio W C Z N	=	03h ? FFh 0 1	;;	(2's com result is r	pleme negati	ent) ive	

SUBWF	Subtract W	from f	
Syntax:	SUBWF f	{,d {,a}}	
Operands:	$0 \leq f \leq 255$		
	$d\in[0,1]$		
	a ∈ [0,1]		
Operation:	$(f) - (W) \rightarrow 0$	dest	
Status Affected:	N, OV, C, D	C, Z	
Encoding:	0101	11da fff	f ffff
Description:	complement result is stor	from register 'f' method). If 'd' ed in W. If 'd' is ck in register 'f'	is '0', the '1', the result
		e Access Banl e BSR is used	
	set is enable in Indexed L mode whene Section 26. Bit-Orientee	d the extended ed, this instruct iteral Offset Ac ever f ≤ 95 (5Fh 2.3 "Byte-Orie d Instructions et Mode" for d	ion operates Idressing n). See nted and in Indexed
Words:	1		
Cycles:	1		
Q Cycle Activity:			
Q1	Q2	Q3	Q4
Decode	Read	Process	Write to
	register 'f'	Data	destination
Example 1:	SUBWF	REG, 1, 0	
Before Instruc			
REG W	= 3 = 2		
C	= ?		
After Instructio			
REG W	= 1 = 2		
Ç		result is positiv	e
Z N	= 0 = 0		
Example 2:	SUBWF	REG, 0, 0	
Before Instruc	tion		
REG W	= 2 = 2		
C	- 2 - ?		
After Instruction	on		
REG W	= 2 = 0		
С	-	result is zero	
Z	= 1 = 0		
Example 3:	SUBWF	REG, 1, 0	
Before Instruc		1110, 1, 0	
REG	= 1		
W C	= 2 = ?		
After Instruction	•		
REG	= FFh ;(2's complement	nt)
W C	= 2 = 0 :	result is negativ	ve
Ž N	= 0 = 1		-

	Subtract	N from f with E	Borrow
Syntax:	SUBWFB	f {,d {,a}}	
Operands:	$0 \le f \le 255$	5	
	d ∈ [0,1]		
	a ∈ [0,1]	_	
Operation:	(f) – (W) –	$(\overline{C}) \rightarrow dest$	
Status Affected:	N, OV, C,	DC, Z	
Encoding:	0101	10da fff	f fff
Description:		V and the Carry	U ()
		ter 'f' (2's comp	
	,	f 'd' is '0', the re is '1', the result	
	in register		
	0	the Access Bar	nk is selected
		the BSR is use	
	GPR bank	-	
	If 'a' is '0'	and the extende	ed instruction
		oled, this instruc	
		Literal Offset A	0
		never f ≤ 95 (5F 6.2.3 "Byte-Ori	,
		ed Instruction	
	Literal Of	fset Mode" for	details.
Words:	1		
Cycles:	1		
Q Cycle Activity:			
Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination
Evenue 4:			destination
Example 1: Before Instruc	SUBWFB	REG, 1, 0	
REG			
	= 19h	(0001 10	01)
W	= 0Dh	(0001 10 (0000 11	
W C	= 0Dh = 1		
W	= 0Dh = 1	(0000 11	
ଜ C After Instructio REG ୴	= 0Dh = 1 on = 0Ch = 0Dh	(0000 11	01) 11)
W C After Instructio REG W C Z	= 0Dh = 1 on = 0Ch	(0000 11) (0000 10 (0000 11)	01) 11) 01)
W C After Instructio REG W C Z N	= 0Dh = 1 on = 0Ch = 0Dh = 1 = 0 = 0	(0000 11) (0000 10) (0000 11) ; result is p	01) 11) 01)
M C After Instructio REG W C Z N N <u>Example 2:</u>	= 0Dh = 1 0n = 0Ch = 0Dh = 1 = 0 = 0 SUBWFB	(0000 11) (0000 10 (0000 11)	01) 11) 01)
M C After Instructio REG W C Z N <u>Example 2:</u> Before Instruc	= 0Dh = 1 on = 0Ch = 0Dh = 1 = 0 = 0 SUBWFB	(0000 11) (0000 10) (0000 11) ; result is po REG, 0, 0	01) 11) 01) psitive
W C After Instructio REG W C Z N N Example 2: Before Instruc REG W	= 0Dh = 1 on = 0Ch = 0Dh = 1 = 0 = 0 SUBWFB tion = 1Bh = 1Ah	(0000 11) (0000 10) (0000 11) ; result is p	01) 11) 01) ositive 11)
M C After Instructio REG W C Z N Example 2: Before Instruc REG W C	= 0Dh = 1 on = 0Ch = 0Ch = 1 = 0 = 0 SUBWFB tion = 1Bh = 1Ah = 0	(0000 11) (0000 10) (0000 11) ; result is pr REG, 0, 0 (0001 10)	01) 11) 01) ositive 11)
M C After Instructio REG W C Z N Example 2: Before Instruct REG W C After Instructio	= 0Dh = 1 on = 0Ch = 0Dh = 1 = 0 = 0 SUBWFB tion = 1Bh = 1Ah = 0 on	(0000 11) (0000 10 (0000 11) ; result is pr REG, 0, 0 (0001 10 (0001 10)	01) 11) 01) ositive 11) 10)
W C After Instructio REG W C Z N Example 2: Before Instruct REG W C After Instructio REG W	= 0Dh = 1 on = 0Ch = 0Dh = 1 = 0 = 0 SUBWFB tion = 1Bh = 1Ah = 0 on = 1Bh = 00h	(0000 11) (0000 10) (0000 11) ; result is pr REG, 0, 0 (0001 10)	01) 11) 01) ositive 11) 10)
W C After Instruction REG W C Z N N Example 2: Before Instruct REG W C After Instruction REG W C	= 0Dh = 1 on = 0Ch = 0Dh = 1 = 0 = 0 SUBWFB tion = 1Bh = 1Ah = 0 on = 1Bh	(0000 11) (0000 10) (0000 11) ; result is pr REG, 0, 0 (0001 10) (0001 10)	01) 11) 01) ositive 11) 10) 11)
M C After Instructio REG W C Z N Example 2: Before Instruct REG W C After Instructio REG W	= 0Dh = 1 on = 0Ch = 0Dh = 1 = 0 = 0 SUBWFB tion = 1Bh = 1Ah = 0 on = 1Bh = 00h = 1	(0000 11) (0000 10 (0000 11) ; result is pr REG, 0, 0 (0001 10 (0001 10)	01) 11) 01) ositive 11) 10) 11)
W C After Instruction REG W Z N Example 2: Before Instruct REG W C After Instruction REG W C Z	= 0Dh = 1 on = 0Ch = 0Dh = 1 = 0 = 0 SUBWFB tion = 1Bh = 0 on = 1Bh = 00h = 1 = 1	(0000 11) (0000 10) (0000 11) ; result is pr REG, 0, 0 (0001 10) (0001 10)	01) 11) 01) ositive 11) 10) 11)
W C After Instruction REG W C Z N Example 2: Before Instruction REG W C After Instruction REG W C After Instruction REG W C After Instruction REG W C After Instruction REG W C S After Instruction REG W C S After Instruction REG W C S S S S S S S S S S S S S S S S S S	= 0Dh = 1 on = 0Ch = 0Dh = 1 = 0 SUBWFB tion = 1Bh = 0 0 = 1Bh = 0 0 = 1 = 1 = 0 SUBWFB tion	(0000 11) (0000 10) (0000 11) ; result is pr REG, 0, 0 (0001 10) (0001 10) (0001 10) ; result is zo REG, 1, 0	01) 11) 01) ositive 11) 10) 11) ero
W C After Instruction REG W C Z N Example 2: Before Instruction REG W C After Instruction REG W C After Instruction REG W C After Instruction REG W C After Instruction REG W C After Instruction REG	= 0Dh = 1 on = 0Ch = 0Dh = 1 = 0 = 0 SUBWFB tion = 1Bh = 0 0 = 1Bh = 0 0 = 1 = 1 = 0 SUBWFB	(0000 11) (0000 10) (0000 11) ; result is pr REG, 0, 0 (0001 10) (0001 10) (0001 10) ; result is ze REG, 1, 0 (0000 00)	<pre>D1) 11) D1) Dsitive 11) 10) 11) Pro 11)</pre>
M C After Instruction REG W C Z N Example 2: Before Instruction REG W C After Instruction REG W C Z N Example 3: Before Instruct Z N Example 3: Before Instruction REG W C	= 0Dh = 1 on = 0Ch = 0Ch = 1 = 0 = 0 SUBWFB tion = 1Bh = 1Ah = 0 on = 1Bh = 1 = 1 = 0 SUBWFB tion = 1 = 1 = 0 = 0 = 1 = 0 = 0 = 0 SUBWFB	(0000 11) (0000 10) (0000 11) ; result is pr REG, 0, 0 (0001 10) (0001 10) (0001 10) ; result is zo REG, 1, 0	<pre>D1) 11) D1) Dsitive 11) 10) 11) Pro 11)</pre>
M C After Instruction REG W C Z N Example 2: Before Instruction REG W C After Instruction REG W C Z N Example 3: Before Instruct REG W C After Instruction REG M C After Instruction REG M C	= 0Dh = 1 on = 0Ch = 0Dh = 1 = 0 = 0 SUBWFB tion = 1Bh = 0 N = 1Ah = 0 ON = 1 = 1 = 0 SUBWFB tion = 0 SUBWFB tion = 0 = 0 SUBWFB	(0000 11) (0000 10 (0000 11) ; result is pr REG, 0, 0 (0001 10 (0001 10) (0001 10) ; result is ze REG, 1, 0 (0000 00 (0000 11)	<pre>D1) 11) D1) Dsitive 11) 10) 11) ero 11) D10</pre>
M C After Instruction REG W C Z N Example 2: Before Instruction REG W C After Instruction REG W C Z N Example 3: Before Instruct Z N Example 3: Before Instruction REG W C	= 0Dh = 1 on = 0Ch = 0Ch = 1 = 0 = 0 SUBWFB tion = 1Bh = 1Ah = 0 on = 1Bh = 1 = 1 = 0 SUBWFB tion = 1 = 1 = 0 = 0 = 1 = 0 = 0 = 0 SUBWFB	(0000 11) (0000 10) (0000 11) ; result is pr REG, 0, 0 (0001 10) (0001 10) (0001 10) ; result is ze REG, 1, 0 (0000 00)	D1) 11) D1) Dsitive 11) 10) 11) ero 11) D1) D1) D0)
W C After Instruction REG W C Z N Example 2: Before Instruction REG W C After Instruction REG W C Z N Example 3: Before Instruction REG W C After Instruction REG W C After Instruction REG W C After Instruction REG	= 0Dh = 1 on = 0Ch = 0 SUBWFB tion = 1Bh = 0 SUBWFB tion = 1Bh = 0 on = 1Bh = 0 SUBWFB tion = 1 = 0 SUBWFB tion = 1 = 0 SUBWFB = 1 = 0 SUBWFB = 1 = 0 SUBWFB = 1 = 0 = 0 SUBWFB = 1 = 0 SUBWFB = 1 = 0 = 1 = 0 SUBWFB = 0	(0000 11) (0000 10) (0000 11) ; result is pr REG, 0, 0 (0001 10) (0001 10) (0001 10) ; result is ze REG, 1, 0 (0000 00) (0000 11) (1111 01)	D1) 11) D1) Dsitive 11) 10) 11) ero 11) D1) D0)
W C After Instruction REG W C Z N Example 2: Before Instruction REG W C After Instruction REG W C Z N Example 3: Before Instruction REG W C After Instruction REG W C After Instruction REG	= 0Dh = 1 0Dh = 0Ch = 0Dh = 1 = 0 SUBWFB tion = 1Bh = 0 SUBWFB tion = 1Bh = 0 SUBWFB tion = 0 SUBWFB tion = 1 = 0 SUBWFB tion = 0 SUBWFB tion = 1 = 0 SUBWFB tion = 1 = 0 SUBWFB tion = 1 = 0 SUBWFB = 1 = 0 SUBWFB = 1 = 0 SUBWFB = 1 = 0 = 0 SUBWFB = 1 = 0 SUBWFB = 1 = 0 = 1 = 0 = 1 = 0 = 1 = 0 SUBWFB = 1 = 0 = 1 = 0 SUBWFB = 1 = 0 SCA = 1 = 1 SUBWFB = 1 SCA = 1 SCA SCA = 1 SCA SCA SCA SCA SCA SCA SCA SCA SCA SCA	(0000 11) (0000 10) (0000 11) ; result is pr REG, 0, 0 (0001 10) (0001 10) (0001 10) ; result is ze REG, 1, 0 (0000 00) (0000 11) ; [2's comp]	D1) 11) D5itive 11) 10) 11) ero 11) D1) D1) D1)

SWAPF	Swap f				
Syntax:	SWAPF f	{,d {,a}}			
Operands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$	d ∈ [0,1]			
Operation:	(f<3:0>) → (f<7:4>) →				
Status Affected:	None				
Encoding:	0011	10da	ffff	ffff	
Description:	'f' are exchais placed in	The upper and lower nibbles of register 'f' are exchanged. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed in register 'f'.			
	If 'a' is '0', t If 'a' is '1', t GPR bank.				
	If 'a' is '0' a set is enabl in Indexed mode wher Section 26 Bit-Oriente Literal Offs	led, this ins Literal Offs never f ≤ 95 .2.3 "Byte- ed Instruct	truction et Addre (5Fh). Oriente ions in	operates essing See ed and Indexed	
Words:	1				
Cycles:	1				
Q Cycle Activity:					
Q1	Q2	Q3		Q4	
Decode	Read register 'f'	Process Data		/rite to stination	
Example:	SWAPF F	REG, 1, 0)		
Before Instruction					
REG	= 53h				
After Instructio REG	on = 35h				

TBLRD *+ ;

Table Read (Continued)

=

=

34h

01A358h

TBL	RD	Table Read					
Synta	ax:	TBLRD (*;	*+; *	-; +*)			
Oper	ands:	None					
•	ands: ation:	None if TBLRD *, (Prog Mem (TBLPTR)) \rightarrow TABLAT; TBLPTR – No Change if TBLRD *+, (Prog Mem (TBLPTR)) \rightarrow TABLAT; (TBLPTR) + 1 \rightarrow TBLPTR if TBLRD *-, (Prog Mem (TBLPTR)) \rightarrow TABLAT; (TBLPTR) – 1 \rightarrow TBLPTR if TBLRD +*, (TBLPTR) + 1 \rightarrow TBLPTR;					
		(Prog Mem				ABLA	Т
Statu	s Affected:	None					
Enco	ding:	0000	00	000	000	00	10nn nn=0 * =1 *+ =2 *- =3 +*
Desc	ription:	This instruct of Program program me Pointer (TBI	Men mor	nory (F y, a po	P.M.).	To ad	dress the
		The TBLPT each byte in has a 2-Mby	the	progra	am me	emory	
		TBLPTR)] = (nt Byte of bry Word
		TBLPTR[()] = 1				it Byte of bry Word
		The TBLRD of TBLPTR				nodify	the value
		 no chang 	е				
		post-incre	emer	nt			
		 post-decr 	eme	ent			
		 pre-increi 	men	t			
Word	ls:	1					
Cycle	Cycles: 2						
QC	Q Cycle Activity:						
	Q1	Q2		C	13		Q4
	Decode	No operation		N opera	-	ор	No eration
	N.L	NI			-	NI	

No operation

(Write

TABLAT)

No

operation

Before Instruction TABLAT TBLPTR MEMORY(00A356h) 55h 00A356h = = = 34h After Instruction TABLAT = 34h TBLPTR = 00A357h TBLRD +* ; Example 2: Before Instruction TABLAT TBLPTR MEMORY(01A357h) MEMORY(01A358h) AAh 01A357h = = = 12h = 34h After Instruction

TABLAT

TBLPTR

TBLRD

Example 1:

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No

operation

No operation

(Read Program

Memory)

(TABLAT) TBLPTR - if TBLWT* (TABLAT) (TBLPTR) if TBLWT*	, → Holdin - No Chan +,	g Register	.,			
if TBLWT* (TABLAT) TBLPTR - if TBLWT* (TABLAT) (TBLPTR) if TBLWT*	→ Holding - No Chan +,		 ,			
(TABLAT) TBLPTR - if TBLWT* (TABLAT) (TBLPTR) if TBLWT*	→ Holding - No Chan +,		., ,			
(TBLPTR) if TBLWT+ (TBLPTR)	if TBLWT*, (TABLAT) \rightarrow Holding Register; TBLPTR – No Change if TBLWT*+, (TABLAT) \rightarrow Holding Register; (TBLPTR) + 1 \rightarrow TBLPTR if TBLWT*-, (TABLAT) \rightarrow Holding Register; (TBLPTR) – 1 \rightarrow TBLPTR if TBLWT+*, (TBLPTR) + 1 \rightarrow TBLPTP:					
	\rightarrow Holding	g Register	•			
	r	r				
0000	0000	0000	11nn nn=0 *			
			=1 *+			
			=2 *-			
			=3 +*			
TBLPTR to determine which of the 8 holding registers the TABLAT is written to. The holding registers are used to program the contents of Program Memory (P.M.). (Refer to Section 6.0 "Memory Organization" for additional details on programming Flash memory.)						
TBLPTR h The LSb c	nas a 2-Mb of the TBL	oyte addre PTR selec	ess range. Its which			
TBLPT		of Prograr	nificant Byte n Memory			
TBLPT		of Prograr	ificant Byte n Memory			
			odify the			
	•					
•	emont					
-						
-						
01	02	03	Q4			
			Q4 No			
Decoue	-	-	operation			
No operation	No	No	No operation (Write to Holding Register)			
	(TBLPTR) (TABLAT) None 0000 This instru TBLPTR t 8 holding i to. The ho program the (P.M.). (Re Organizat programm The TBLPT the TBLPTR t THE LSb of byte of the access. TBLPT TBLPT TBLPT TBLPT TBLPT TBLPT TBLPT TBLPT 1 0 post-inc 0 post-inc 1 2 Q1 Decode No	$\begin{array}{c c c c c c c c c c c c c c c c c c c $	$\begin{array}{c c c c c c c c c c c c c c c c c c c $			

TBLWT Table Write (Continued)

Example 1:	TBLWT *+;		
Before Instru	ction		
		= =	55h 00A356h
(00A356		=	FFh
After Instructi	ons (table write	compl	letion)
TABLAT		=	55h
	R IG REGISTER	=	00A357h
(00A356		=	55h
Example 2:	TBLWT +*;		
Before Instru	ction		
TABLAT		=	34h
	R IG REGISTER	=	01389Ah
(01389A		=	FFh
(01389E	Bh)	=	FFh
After Instructi	on (table write c	omple	tion)
TABLAT		=	34h
	R IG REGISTER	=	01389Bh
(01389A		=	FFh
(01389E		=	34h

TSTF	sz	Test f, Skip	if 0		
Synta	ax:	TSTFSZ f {	,a}		
Oper	ands:	0 ≤ f ≤ 255 a ∈ [0,1]			
Oper	ation:	skip if f = 0			
Statu	s Affected:	None			
Enco	ding:	0110	011a fff	f ffff	
Desc	ription:	during the c is discarded	e next instructio current instruct d and a NOP is a two-cycle in	ion execution executed,	
		If 'a' is '0', the Access Bank is selected. If 'a' is '⊥', the BSR is used to select the GPR bank.			
		If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.			
Word	ls:	1			
Cycle	es:		rcles if skip an a 2-word instru		
QC	ycle Activity:			.	
	Q1 Decode	Q2 Read	Q3 Process	Q4 No	
	Decode	register 'f'	Data	operation	
lf sk	ip:	<u> </u>			
	Q1	Q2	Q3	Q4	
	No	No	No	No	
16 - 1-	operation	operation	operation	operation	
IT SK	Ip and followed	d by 2-word in: Q2		01	
	No	No	Q3 No	Q4 No	
	operation	operation	operation	operation	
	No	No	No	No	
	operation	operation	operation	operation	
<u>Exan</u>	nple:	HERE T NZERO : ZERO :		, 1	
	Before Instruc	tion			
	PC = Address (HERE)				
	After Instructic If CNT	on = 00	h		
	PC		n, dress (ZERO)	
	If CNT PC	≠ 00 = Ad	h, dress (NZERO))	

XOR	LW	Exclusive OR Literal with W				
Synta	ax:	XORLW	XORLW k			
Oper	ands:	$0 \le k \le 25$	5			
Oper	ation:	(W) .XOR	$k \rightarrow W$			
Statu	is Affected:	N, Z				
Enco	oding:	0000	1010	kkkl	k	kkkk
Desc	ription:	The contents of W are XORed with the 8-bit literal 'k'. The result is placed in W.				
Word	ls:	1				
Cycle	es:	1				
QC	ycle Activity:					
	Q1	Q2	Q3			Q4
	Decode	Read literal 'k'	Proce: Data		W	′rite to W
Exan	nple:	XORLW	0AFh			
Before Instruction W = B5h After Instruction						

1Ah

=

W

XORWF	Exclusive	Exclusive OR W with f			
Syntax:	XORWF	f {,d {,a}}			
Operands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$				
Operation:	(W) .XOR.	(f) \rightarrow dest			
Status Affected:	N, Z				
Encoding:	0001	10da ffi	ff ffff		
Description:	register 'f'. I	DR the content f 'd' is '0', the r s '1', the result ter 'f'.	esult is stored		
		he Access Bar he BSR is use			
	set is enabl in Indexed mode when Section 26	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \le 95$ (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed			
Words:	1				
Cycles:	1				
Q Cycle Activity:					
Q1	Q2	Q3	Q4		
Decode	Read register 'f'	Process Data	Write to destination		
Example: Before Instruc REG W	tion = AFh = B5h	REG, 1, 0			
After Instructio REG W	on = 1Ah = B5h				

26.2 Extended Instruction Set

In addition to the standard 75 instructions of the PIC18 instruction set, the PIC18F87J90 family family of devices also provide an optional extension to the core CPU functionality. The added features include eight additional instructions that augment Indirect and Indexed Addressing operations and the implementation of Indexed Literal Offset Addressing for many of the standard PIC18 instructions.

The additional features of the extended instruction set are enabled by default on unprogrammed devices. Users must properly set or clear the XINST Configuration bit during programming to enable or disable these features.

The instructions in the extended set can all be classified as literal operations, which either manipulate the File Select Registers, or use them for Indexed Addressing. Two of the instructions, ADDFSR and SUBFSR, each have an additional special instantiation for using FSR2. These versions (ADDULNK and SUBULNK) allow for automatic return after execution.

The extended instructions are specifically implemented to optimize re-entrant program code (that is, code that is recursive or that uses a software stack) written in high-level languages, particularly C. Among other things, they allow users working in high-level languages to perform certain operations on data structures more efficiently. These include:

- Dynamic allocation and deallocation of software stack space when entering and leaving subroutines
- Function Pointer invocation
- Software Stack Pointer manipulation
- Manipulation of variables located in a software stack

A summary of the instructions in the extended instruction set is provided in Table 26-3. Detailed descriptions are provided in **Section 26.2.2 "Extended Instruction Set"**. The opcode field descriptions in Table 26-1 (page 340) apply to both the standard and extended PIC18 instruction sets.

Note: The instruction set extension and the Indexed Literal Offset Addressing mode were designed for optimizing applications written in C; the user may likely never use these instructions directly in assembler. The syntax for these commands is provided as a reference for users who may be reviewing code that has been generated by a compiler.

26.2.1 EXTENDED INSTRUCTION SYNTAX

Most of the extended instructions use indexed arguments, using one of the File Select Registers and some offset to specify a source or destination register. When an argument for an instruction serves as part of Indexed Addressing, it is enclosed in square brackets ("[]"). This is done to indicate that the argument is used as an index or offset. The MPASM[™] Assembler will flag an error if it determines that an index or offset value is not bracketed.

When the extended instruction set is enabled, brackets are also used to indicate index arguments in byte-oriented and bit-oriented instructions. This is in addition to other changes in their syntax. For more details, see Section 26.2.3.1 "Extended Instruction Syntax with Standard PIC18 Commands".

Note: In the past, square brackets have been used to denote optional arguments in the PIC18 and earlier instruction sets. In this text and going forward, optional arguments are denoted by braces ("{ }").

Mnemonic,		Description	Cycles	16-Bit Instruction Word				Status
Opera	nds	Description	Cycles	MSb			LSb	Affected
ADDFSR	f, k	Add Literal to FSR	1	1110	1000	ffkk	kkkk	None
ADDULNK	k	Add Literal to FSR2 and Return	2	1110	1000	11kk	kkkk	None
CALLW		Call Subroutine using WREG	2	0000	0000	0001	0100	None
MOVSF	z _s , f _d	Move z _s (source) to 1st word	2	1110	1011	0zzz	ZZZZ	None
		f _d (destination) 2nd word		1111	ffff	ffff	ffff	
MOVSS	z _s , z _d	Move z _s (source) to 1st word	2	1110	1011	lzzz	ZZZZ	None
		z _d (destination) 2nd word		1111	xxxx	XZZZ	ZZZZ	
PUSHL	k	Store Literal at FSR2,	1	1110	1010	kkkk	kkkk	None
		Decrement FSR2						
SUBFSR	f, k	Subtract Literal from FSR	1	1110	1001	ffkk	kkkk	None
SUBULNK	k	Subtract Literal from FSR2 and Return	2	1110	1001	11kk	kkkk	None

TABLE 26-3: EXTENSIONS TO THE PIC18 INSTRUCTION SET

26.2.2 EXTENDED INSTRUCTION SET

ADD	FSR	Add Literal to FSR				
Synta	ax:	ADDFSR	f, k			
Oper	ands:	$0 \le k \le 63$				
		f ∈ [0, 1,	2]			
Oper	ation:	FSR(f) + I	$s \rightarrow FSR($	f)		
Statu	s Affected:	None				
Encoding: 1110 1000 ffkk kkk					kkkk	
Desc	ription:	The 6-bit	iteral 'k' i	s adde	ed to the	
		contents of	of the FSF	R spec	ified by 'f'.	
Word	ls:	1				
Cycle	es:	1				
QC	ycle Activity:					
	Q1	Q2	Q3		Q4	
	Decode	Read	Proces	ss	Write to	
		literal 'k'	Data		FSR	

Example:	ADDFSR	2,	23h
----------	--------	----	-----

Before Instru	ction	
FSR2	=	03FFh
After Instruct		
FSR2	=	0422h

ADD	ULNK	Add Literal to FSR2 and Return				
Synta	ax:	ADDULN	ADDULNK k			
Oper	ands:	$0 \le k \le 63$	$0 \le k \le 63$			
Oper	ation:	FSR2 + k	\rightarrow FSR2	,		
		$(TOS) \rightarrow I$	С			
Statu	s Affected:	None				
Enco	ding:	1110	1000	11kl	k kkkk	
Desc	ription:	contents o	The 6-bit literal 'k' is added to the contents of FSR2. A RETURN is then executed by loading the PC with the TOS.			
		The instru execute; a the second	NOP is p		o cycles to ned during	
		case of the	e ADDFSI 3 (binary	ર instr	is a special ruction, it operates	
Word	ls:	1				
Cycle	es:	2				
QC	ycle Activity:					
	Q1	Q2	Q3		Q4	
	Decode	Read	Proces	SS	Write to	
		literal 'k'	Data		FSR	
	No	No	No		No	
	Operation	Operation	Operat	ion	Operation	
<u>Exan</u>	nple:	ADDULNK 2	23h			

<u>ampie:</u>	A	DDULNK	2
Before Instruct	ion		
FSR2	=	03FFh	
PC	=	0100h	
After Instructio	n		
FSR2	=	0422h	
PC	=	(TOS)	

Note: All PIC18 instructions may take an optional label argument preceding the instruction mnemonic for use in symbolic addressing. If a label is used, the instruction format then becomes: {label} instruction argument(s).

CALI	LW	Subroutine Call Using WREG				
Synta	ax:	CALLW				
Oper	ands:	None				
Oper	ation:	(PC + 2) → (W) → PCL (PCLATH) - (PCLATU) -	, → PCH,			
Statu	s Affected:	None				
Enco	ding:	0000 0000 0001 0100				
Desc	ription	First, the return address (PC + 2) is pushed onto the return stack. Next, the contents of W are written to PCL; the existing value is discarded. Then, the contents of PCLATH and PCLATU are latched into PCH and PCU, respectively. The second cycle is executed as a NOP instruction while the new next instruction is fetched. Unlike CALL, there is no option to update W, STATUS or BSR.				
Word	lo.	1	51A105 01 B5	Π.		
Word		•				
Cycle		2				
QU	ycle Activity: Q1	Q2	Q3	Q4		
ĺ	Decode	Read	Push PC to	No		
	Debbuc	WREG	stack	operation		
	No	No	No	No		
	operation	operation	operation	operation		
Example:HERECALLWBefore Instruction $PC = address (HERE)$ $PCLATH = 10h$ $PCLATU = 00h$ $W = 06h$ After Instruction $PC = 001006h$ $TOS = address (HERE + 2)$ $PCLATH = 10h$ $PCLATU = 00h$						
	W	= 06h				

MOV	'SF	Move Inde	xed to f			
Synta	ax:	MOVSF [z _s], f _d			
Oper	ands:	$0 \le z_s \le 12$ $0 \le f_d \le 409$				
Oper	ation:	((FSR2) +	$z_s) \rightarrow f_d$			
Statu	is Affected:	None				
1st w	oding: vord (source) word (destin.)	1110 1111	1011 ffff	0zz fff		zzzz _s ffff _d
Desc	cription:	moved to c actual addu determined offset ' z_s ', of FSR2. T register is a 'f _d ' in the s can be any space (000 The MOVSH PCL, TOSH destination If the result	The contents of the source register are moved to destination register 'f _d '. The actual address of the source register is determined by adding the 7-bit literal offset 'z _s ', in the first word, to the value of FSR2. The address of the destination register is specified by the 12-bit literal 'f _d ' in the second word. Both addresses can be anywhere in the 4096-byte data space (000h to FFFh). The MOVSF instruction cannot use the PCL, TOSU, TOSH or TOSL as the destination register. If the resultant source address points to an Indirect Addressing register, the			
Word	10.	value retur	ned will be	e uun	-	
Cycle		2				
	ycle Activity:	2				
QU	Q1	Q2	Q3			Q4
	Decode	Determine source addr	Determ source a			Read irce reg
	Decode	No operation No dummy read	No operati	on	reg	Vrite jister 'f' dest)
<u>Exan</u>	n <u>ple:</u> Before Instruc FSR2 Contents	= 80	[05h], 3	REG2		
	of 85h REG2 After Instructio FSR2 Contents	= 33 = 11 on = 80	h			
	of 85h REG2	= 33 = 33				

MOVSS	Move Indexed to Indexed			
Syntax:	MOVSS [z _s], [z _d]			
Operands:	$\begin{array}{l} 0 \leq z_s \leq 127 \\ 0 \leq z_d \leq 127 \end{array}$			
Operation:	((FSR2) +	$z_s) \rightarrow ((F$	SR2) + z _d)
Status Affected:	None			
Encoding: 1st word (source) 2nd word (dest.) Description	$ \begin{array}{ c c c c c } \hline 1110 & 1011 & 1zzz & zzzz_s \\ \hline 1111 & xxxx & xzzz & zzzz_d \\ \hline \\ $			
	addresses of the source and destination registers are determined by adding the 7-bit literal offsets, ' z_s ' or ' z_d ', respectively, to the value of FSR2. Both registers can be located anywhere in the 4096-byte data memory space (000h to FFFh).			
	The MOVSS instruction cannot use the PCL, TOSU, TOSH or TOSL as the destination register.			
	If the resultant source address points to an Indirect Addressing register, the value returned will be 00h. If the resultant destination address points to an Indirect Addressing register, the instruction will execute as a NOP.			
Words:	2			
Cycles:	2			
Q Cycle Activity:				

Q1	Q2	Q3	Q4
Decode	Determine	Determine	Read
	source addr	source addr	source reg
Decode	Determine	Determine	Write
	dest addr	dest addr	to dest reg

Example: MOVSS [05h], [06h]	Example:	MOVSS	[05h],	[06h]
-----------------------------	----------	-------	--------	-------

Before Instruction		
FSR2	=	80h
Contents of 85h Contents	=	33h
of 86h	=	11h
After Instruction		
FSR2	=	80h
Contents of 85h Contents	=	33h
of 86h	=	33h

PUSHL	Store Literal at FSR2, Decrement FSR2			
Syntax:	PUSHL k			
Operands:	$0 \leq k \leq 255$			
Operation:	$k \rightarrow (FSR2)$ FSR2 – 1 –	,		
Status Affected:	None			
Encoding:	1111	1010	kkkk	kkkk
Description:	The 8-bit literal 'k' is written to the data memory address specified by FSR2. FSR2 is decremented by 1 after the operation. This instruction allows users to push			
	values onto			•
Words:	1			
Cycles:	1			
Q Cycle Activity:				
Q1	Q2	C	3	Q4
Decode	Read 'k'	Proc da		Write to destination
Example:	PUSHL 0	8h		

Before Instruction FSR2H:FSR2L = 01ECh Memory (01ECh) = 00h After Instruction FSR2H:FSR2L = 01EBh Memory (01ECh) = 08h

SUB	FSR	Subtract Literal from FSR				
Synta	ax:	SUBFSR	f, k			
Oper	ands:	$0 \le k \le 63$	$0 \le k \le 63$			
		f ∈ [0, 1,	f ∈ [0, 1, 2]			
Oper	ation:	$FSRf - k \rightarrow FSRf$				
Statu	s Affected:	None				
Enco	oding:	1110	1110 1001 ffkk kkkk			
Desc	ription:	The 6-bit	The 6-bit literal 'k' is subtracted from			
		the contents of the FSR specified				
		by 'f'.				
Word	ls:	1				
Cycle	es:	1				
QC	ycle Activity:					
	Q1	Q2	Q3		Q4	
	Decode	Read	Proce	SS	Write to	
		register 'f' Data destination				
<u>Exan</u>	Example: SUBFSR 2, 23h					

	Decoue		louu	11000000	l
		reg	jister 'f'	Data	
Exam	<u>nple:</u>	S	UBFSR 2	2, 23h	
	Before Instruc	tion			
	FSR2	=	03FFh		
	After Instructio	n			
	FSR2	=	03DCh		

SUB	ULNK	Subtract Literal from FSR2 and Return					
Synta	ax:	SUBULNK k					
Oper	ands:	$0 \le k \le 63$					
Oper	ation:	$FSR2 - k + (TOS) \rightarrow F$		SR2,			
Statu	s Affected:	None					
Enco	ding:	1110	10	01	11kk		kkkk
Desc	ription:	The 6-bit literal 'k' is subtracted from the contents of the FSR2. A RETURN is then executed by loading the PC with the TOS.					
		The instruction takes two cycles to execute; a NOP is performed during the second cycle.					
		This may be thought of as a special case of the SUBFSR instruction, where f = 3 (binary '11'); it operates only on FSR2.					
Word	ls:	1					
Cycle	es:	2					
QC	ycle Activity:						
	Q1	Q2		C	23		Q4
	Decode	Read register	ʻf'		cess ata		Write to estination
	No Operation	No Operatio	n		o ation	0	No peration

Example: SUBULNK 23h

Before Instruction				
FSR2	=	03FFh		
PC	=	0100h		
After Instructi	on			
FSR2	=	03DCh		
PC	=	(TOS)		
	=			

26.2.3 BYTE-ORIENTED AND BIT-ORIENTED INSTRUCTIONS IN INDEXED LITERAL OFFSET MODE

Note:	Enabling the PIC18 instruction set exten-				
	sion may cause legacy applications to				
	behave erratically or fail entirely.				

In addition to eight new commands in the extended set, enabling the extended instruction set also enables Indexed Literal Offset Addressing (Section 6.6.1 "Indexed Addressing with Literal Offset"). This has a significant impact on the way that many commands of the standard PIC18 instruction set are interpreted.

When the extended set is disabled, addresses embedded in opcodes are treated as literal memory locations: either as a location in the Access Bank (a = 0) or in a GPR bank designated by the BSR (a = 1). When the extended instruction set is enabled and a = 0, however, a file register argument of 5Fh or less is interpreted as an offset from the pointer value in FSR2 and not as a literal address. For practical purposes, this means that all instructions that use the Access RAM bit as an argument – that is, all byte-oriented and bit-oriented instructions, or almost half of the core PIC18 instructions – may behave differently when the extended instruction set is enabled.

When the content of FSR2 is 00h, the boundaries of the Access RAM are essentially remapped to their original values. This may be useful in creating backward-compatible code. If this technique is used, it may be necessary to save the value of FSR2 and restore it when moving back and forth between C and assembly routines in order to preserve the Stack Pointer. Users must also keep in mind the syntax requirements of the extended instruction set (see Section 26.2.3.1 "Extended Instruction Syntax with Standard PIC18 Commands").

Although the Indexed Literal Offset mode can be very useful for dynamic stack and pointer manipulation, it can also be very annoying if a simple arithmetic operation is carried out on the wrong register. Users who are accustomed to the PIC18 programming must keep in mind that, when the extended instruction set is enabled, register addresses of 5Fh or less are used for Indexed Literal Offset Addressing.

Representative examples of typical byte-oriented and bit-oriented instructions in the Indexed Literal Offset mode are provided on the following page to show how execution is affected. The operand conditions shown in the examples are applicable to all instructions of these types.

26.2.3.1 Extended Instruction Syntax with Standard PIC18 Commands

When the extended instruction set is enabled, the file register argument, 'f', in the standard byte-oriented and bit-oriented commands is replaced with the literal offset value, 'k'. As already noted, this occurs only when 'f' is less than or equal to 5Fh. When an offset value is used, it must be indicated by square brackets ("[]"). As with the extended instructions, the use of brackets indicates to the compiler that the value is to be interpreted as an index or an offset. Omitting the brackets, or using a value greater than 5Fh within the brackets, will generate an error in the MPASM Assembler.

If the index argument is properly bracketed for Indexed Literal Offset Addressing, the Access RAM argument is never specified; it will automatically be assumed to be '0'. This is in contrast to standard operation (extended instruction set disabled), when 'a' is set on the basis of the target address. Declaring the Access RAM bit in this mode will also generate an error in the MPASM Assembler.

The destination argument 'd' functions as before.

In the latest versions of the MPASM Assembler, language support for the extended instruction set must be explicitly invoked. This is done with either the command line option, $/_{Y}$, or the PE directive in the source listing.

26.2.4 CONSIDERATIONS WHEN ENABLING THE EXTENDED INSTRUCTION SET

It is important to note that the extensions to the instruction set may not be beneficial to all users. In particular, users who are not writing code that uses a software stack may not benefit from using the extensions to the instruction set.

Additionally, the Indexed Literal Offset Addressing mode may create issues with legacy applications written to the PIC18 assembler. This is because instructions in the legacy code may attempt to address registers in the Access Bank below 5Fh. Since these addresses are interpreted as literal offsets to FSR2 when the instruction set extension is enabled, the application may read or write to the wrong data addresses.

When porting an application to the PIC18F87J90 family family, it is very important to consider the type of code. A large, re-entrant application that is written in C and would benefit from efficient compilation will do well when using the instruction set extensions. Legacy applications that heavily use the Access Bank will most likely not benefit from using the extended instruction set.

ADDWF	ADD W to (Indexed I		fset mo	de)	
Syntax:	ADDWF	[k] {,d}			
Operands:	$\begin{array}{l} 0 \leq k \leq 95 \\ d \in [0,1] \end{array}$				
Operation:	(W) + ((FS	R2) + k) -	→ dest		
Status Affected:	N, OV, C,	DC, Z			
Encoding:	0010	01d0	kkkk	kkkk	
Description:	The contents of W are added to the contents of the register indicated by FSR2, offset by the value 'k'.				
	If 'd' is '0', is '1', the r register 'f'.	esult is st		d in W. If 'd' ck in	
Words:	1				
Cycles:	1				
Q Cycle Activity:					
Q1	Q2	Q3		Q4	
Decode	Read 'k'	Proce Data		Write to estination	
Example:	ADDWF	[OFST]	, 0		
Before Instruction W OFST FSR2 Contents of 0A2Ch After Instruction W Contents of 0A2Ch	= = = =	17h 2Ch 0A00r 20h 37h 20h	1		

BSF	Bit Set Ind (Indexed L	exed iteral Offset r	node)			
Syntax:	BSF [k], b					
Operands:	$\begin{array}{l} 0 \leq f \leq 95 \\ 0 \leq b \leq 7 \end{array}$					
Operation:	$1 \rightarrow ((FSR))$	2) + k) 				
Status Affected:	None					
Encoding:	1000	bbb0 kkl	kk kkkk			
Description:		e register indica e value 'k', is s				
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3	Q4			
Decode	Read register 'f'	Process Data	Write to destination			
Example:		FLAG_OFST]	, 7			
Before Instructi FLAG_OF FSR2		0Ah 0A00h				
Contents of 0A0Ah After Instructior	=	55h				
Contents of 0A0Ah	=	D5h				
SETF	Set Indexe (Indexed L	d iteral Offset r	node)			
-	(Indexed L		node)			
Syntax:			node)			
Syntax: Operands:	(Indexed L SETF [k] $0 \le k \le 95$	iteral Offset r	node)			
Syntax:	(Indexed L SETF [k]	iteral Offset r	node)			
Syntax: Operands: Operation: Status Affected:	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS	iteral Offset r				
Syntax: Operands: Operation:	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The conten	SR2) + k)	kk kkkk er indicated by			
Syntax: Operands: Operation: Status Affected: Encoding:	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The conten	iteral Offset (SR2) + k) 1000 kk	kk kkkk er indicated by			
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words:	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The conten FSR2, offset	SR2) + k)	kk kkkk er indicated by			
Syntax: Operands: Operation: Status Affected: Encoding: Description:	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The conten FSR2, offse 1	SR2) + k)	kk kkkk er indicated by			
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The conten FSR2, offse 1 1 2	SR2) + k) 1000 kk ts of the regist et by 'k', are se	kk kkkk er indicated by et to FFh. Q4			
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity:	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The conten FSR2, offse 1 1	SR2) + k) 1000 kk ts of the regist et by 'k', are se	kk kkkk er indicated by et to FFh.			
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The conten FSR2, offse 1 1 Q2 Read 'k'	SR2) + k) 1000 kk ts of the regist et by 'k', are se Q3 Process	kk kkkk er indicated by et to FFh. Q4 Write			
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode Example: Before Instructi	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The conten FSR2, offse 1 1 2 Read 'k' SETF [on	GR2) + k) 1000 kk ts of the regist et by 'k', are se Q3 Process Data COFST]	kk kkkk er indicated by et to FFh. Q4 Write			
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode Example: Before Instructi OFST FSR2	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The conten FSR2, offse 1 1 2 Read 'k' SETF [on = 20	GR2) + k) 1000 kk ts of the regist et by 'k', are se Q3 Process Data COFST]	kk kkkk er indicated by et to FFh. Q4 Write			
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode Example: Before Instructi OFST FSR2 Contents	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The conten FSR2, offset 1 1 Q2 Read 'k' SETF [on = 2C = 0A	SR2) + k) 1000 kk: ts of the regist et by 'k', are se Q3 Process Data CoFST] Ch 000h	kk kkkk er indicated by et to FFh. Q4 Write			
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode Example: Before Instructi OFST FSR2	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The conten FSR2, offset 1 1 Q2 Read 'k' SETF [on = 20 = 00	SR2) + k) 1000 kk: ts of the regist et by 'k', are se Q3 Process Data CoFST] Ch 000h	kk kkkk er indicated by et to FFh. Q4 Write			
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode Example: Before Instructi OFST FSR2 Contents of 0A2Ch	(Indexed L SETF [k] $0 \le k \le 95$ FFh \rightarrow ((FS None 0110 The conten FSR2, offset 1 1 Q2 Read 'k' SETF [on = 20 = 00	iteral Offset r SR2) + k) 1000 kk ts of the regist et by 'k', are se Q3 Process Data OFST] Ch 000h h	kk kkkk er indicated by et to FFh. Q4 Write			

26.2.5 SPECIAL CONSIDERATIONS WITH MICROCHIP MPLAB[®] IDE TOOLS

The latest versions of Microchip's software tools have been designed to fully support the extended instruction set for the PIC18F87J90 family family. This includes the MPLAB C18 C Compiler, MPASM assembly language and MPLAB Integrated Development Environment (IDE).

When selecting a target device for software development, MPLAB IDE will automatically set default Configuration bits for that device. The default setting for the XINST Configuration bit is '1', enabling the extended instruction set and Indexed Literal Offset Addressing. For proper execution of applications developed to take advantage of the extended instruction set, XINST must be set during programming.

To develop software for the extended instruction set, the user must enable support for the instructions and the Indexed Addressing mode in their language tool(s). Depending on the environment being used, this may be done in several ways:

- A menu option or dialog box within the environment that allows the user to configure the language tool and its settings for the project
- A command line option
- · A directive in the source code

These options vary between different compilers, assemblers and development environments. Users are encouraged to review the documentation accompanying their development systems for the appropriate information.

27.0 DEVELOPMENT SUPPORT

The PIC[®] microcontrollers and dsPIC[®] digital signal controllers are supported with a full range of software and hardware development tools:

- Integrated Development Environment
- MPLAB[®] IDE Software
- Compilers/Assemblers/Linkers
 - MPLAB C Compiler for Various Device Families
 - HI-TECH C for Various Device Families
 - MPASM[™] Assembler
 - MPLINK[™] Object Linker/ MPLIB[™] Object Librarian
 - MPLAB Assembler/Linker/Librarian for Various Device Families
- Simulators
 - MPLAB SIM Software Simulator
- Emulators
 - MPLAB REAL ICE™ In-Circuit Emulator
- In-Circuit Debuggers
 - MPLAB ICD 3
 - PICkit™ 3 Debug Express
- Device Programmers
 - PICkit[™] 2 Programmer
 - MPLAB PM3 Device Programmer
- Low-Cost Demonstration/Development Boards, Evaluation Kits, and Starter Kits

27.1 MPLAB Integrated Development Environment Software

The MPLAB IDE software brings an ease of software development previously unseen in the 8/16/32-bit microcontroller market. The MPLAB IDE is a Windows[®] operating system-based application that contains:

- A single graphical interface to all debugging tools
 - Simulator
 - Programmer (sold separately)
 - In-Circuit Emulator (sold separately)
 - In-Circuit Debugger (sold separately)
- A full-featured editor with color-coded context
- A multiple project manager
- Customizable data windows with direct edit of contents
- High-level source code debugging
- Mouse over variable inspection
- Drag and drop variables from source to watch windows
- · Extensive on-line help
- Integration of select third party tools, such as IAR C Compilers

The MPLAB IDE allows you to:

- Edit your source files (either C or assembly)
- One-touch compile or assemble, and download to emulator and simulator tools (automatically updates all project information)
- Debug using:
 - Source files (C or assembly)
 - Mixed C and assembly
 - Machine code

MPLAB IDE supports multiple debugging tools in a single development paradigm, from the cost-effective simulators, through low-cost in-circuit debuggers, to full-featured emulators. This eliminates the learning curve when upgrading to tools with increased flexibility and power.

27.2 MPLAB C Compilers for Various Device Families

The MPLAB C Compiler code development systems are complete ANSI C compilers for Microchip's PIC18, PIC24 and PIC32 families of microcontrollers and the dsPIC30 and dsPIC33 families of digital signal controllers. These compilers provide powerful integration capabilities, superior code optimization and ease of use.

For easy source level debugging, the compilers provide symbol information that is optimized to the MPLAB IDE debugger.

27.3 HI-TECH C for Various Device Families

The HI-TECH C Compiler code development systems are complete ANSI C compilers for Microchip's PIC family of microcontrollers and the dsPIC family of digital signal controllers. These compilers provide powerful integration capabilities, omniscient code generation and ease of use.

For easy source level debugging, the compilers provide symbol information that is optimized to the MPLAB IDE debugger.

The compilers include a macro assembler, linker, preprocessor, and one-step driver, and can run on multiple platforms.

27.4 MPASM Assembler

The MPASM Assembler is a full-featured, universal macro assembler for PIC10/12/16/18 MCUs.

The MPASM Assembler generates relocatable object files for the MPLINK Object Linker, Intel[®] standard HEX files, MAP files to detail memory usage and symbol reference, absolute LST files that contain source lines and generated machine code and COFF files for debugging.

The MPASM Assembler features include:

- · Integration into MPLAB IDE projects
- User-defined macros to streamline assembly code
- Conditional assembly for multi-purpose source files
- Directives that allow complete control over the assembly process

27.5 MPLINK Object Linker/ MPLIB Object Librarian

The MPLINK Object Linker combines relocatable objects created by the MPASM Assembler and the MPLAB C18 C Compiler. It can link relocatable objects from precompiled libraries, using directives from a linker script.

The MPLIB Object Librarian manages the creation and modification of library files of precompiled code. When a routine from a library is called from a source file, only the modules that contain that routine will be linked in with the application. This allows large libraries to be used efficiently in many different applications.

The object linker/library features include:

- Efficient linking of single libraries instead of many smaller files
- Enhanced code maintainability by grouping related modules together
- Flexible creation of libraries with easy module listing, replacement, deletion and extraction

27.6 MPLAB Assembler, Linker and Librarian for Various Device Families

MPLAB Assembler produces relocatable machine code from symbolic assembly language for PIC24, PIC32 and dsPIC devices. MPLAB C Compiler uses the assembler to produce its object file. The assembler generates relocatable object files that can then be archived or linked with other relocatable object files and archives to create an executable file. Notable features of the assembler include:

- · Support for the entire device instruction set
- · Support for fixed-point and floating-point data
- Command line interface
- · Rich directive set
- Flexible macro language
- · MPLAB IDE compatibility

27.7 MPLAB SIM Software Simulator

The MPLAB SIM Software Simulator allows code development in a PC-hosted environment by simulating the PIC MCUs and dsPIC[®] DSCs on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a comprehensive stimulus controller. Registers can be logged to files for further run-time analysis. The trace buffer and logic analyzer display extend the power of the simulator to record and track program execution, actions on I/O, most peripherals and internal registers.

The MPLAB SIM Software Simulator fully supports symbolic debugging using the MPLAB C Compilers, and the MPASM and MPLAB Assemblers. The software simulator offers the flexibility to develop and debug code outside of the hardware laboratory environment, making it an excellent, economical software development tool.

27.8 MPLAB REAL ICE In-Circuit Emulator System

MPLAB REAL ICE In-Circuit Emulator System is Microchip's next generation high-speed emulator for Microchip Flash DSC and MCU devices. It debugs and programs PIC[®] Flash MCUs and dsPIC[®] Flash DSCs with the easy-to-use, powerful graphical user interface of the MPLAB Integrated Development Environment (IDE), included with each kit.

The emulator is connected to the design engineer's PC using a high-speed USB 2.0 interface and is connected to the target with either a connector compatible with incircuit debugger systems (RJ11) or with the new high-speed, noise tolerant, Low-Voltage Differential Signal (LVDS) interconnection (CAT5).

The emulator is field upgradable through future firmware downloads in MPLAB IDE. In upcoming releases of MPLAB IDE, new devices will be supported, and new features will be added. MPLAB REAL ICE offers significant advantages over competitive emulators including low-cost, full-speed emulation, run-time variable watches, trace analysis, complex breakpoints, a ruggedized probe interface and long (up to three meters) interconnection cables.

27.9 MPLAB ICD 3 In-Circuit Debugger System

MPLAB ICD 3 In-Circuit Debugger System is Microchip's most cost effective high-speed hardware debugger/programmer for Microchip Flash Digital Signal Controller (DSC) and microcontroller (MCU) devices. It debugs and programs PIC[®] Flash microcontrollers and dsPIC[®] DSCs with the powerful, yet easyto-use graphical user interface of MPLAB Integrated Development Environment (IDE).

The MPLAB ICD 3 In-Circuit Debugger probe is connected to the design engineer's PC using a high-speed USB 2.0 interface and is connected to the target with a connector compatible with the MPLAB ICD 2 or MPLAB REAL ICE systems (RJ-11). MPLAB ICD 3 supports all MPLAB ICD 2 headers.

27.10 PICkit 3 In-Circuit Debugger/ Programmer and PICkit 3 Debug Express

The MPLAB PICkit 3 allows debugging and programming of PIC[®] and dsPIC[®] Flash microcontrollers at a most affordable price point using the powerful graphical user interface of the MPLAB Integrated Development Environment (IDE). The MPLAB PICkit 3 is connected to the design engineer's PC using a full speed USB interface and can be connected to the target via an Microchip debug (RJ-11) connector (compatible with MPLAB ICD 3 and MPLAB REAL ICE). The connector uses two device I/O pins and the reset line to implement in-circuit debugging and In-Circuit Serial Programming[™].

The PICkit 3 Debug Express include the PICkit 3, demo board and microcontroller, hookup cables and CDROM with user's guide, lessons, tutorial, compiler and MPLAB IDE software.

27.11 PICkit 2 Development Programmer/Debugger and PICkit 2 Debug Express

The PICkit[™] 2 Development Programmer/Debugger is a low-cost development tool with an easy to use interface for programming and debugging Microchip's Flash families of microcontrollers. The full featured Windows[®] programming interface supports baseline (PIC10F, PIC12F5xx, PIC16F5xx), midrange (PIC12F6xx, PIC16F), PIC18F, PIC24, dsPIC30, dsPIC33, and PIC32 families of 8-bit, 16-bit, and 32-bit microcontrollers, and many Microchip Serial EEPROM products. With Microchip's powerful MPLAB Integrated Development Environment (IDE) the PICkit[™] 2 enables in-circuit debugging on most PIC[®] microcontrollers. In-Circuit-Debugging runs, halts and single steps the program while the PIC microcontroller is embedded in the application. When halted at a breakpoint, the file registers can be examined and modified.

The PICkit 2 Debug Express include the PICkit 2, demo board and microcontroller, hookup cables and CDROM with user's guide, lessons, tutorial, compiler and MPLAB IDE software.

27.12 MPLAB PM3 Device Programmer

The MPLAB PM3 Device Programmer is a universal, CE compliant device programmer with programmable voltage verification at VDDMIN and VDDMAX for maximum reliability. It features a large LCD display (128 x 64) for menus and error messages and a modular, detachable socket assembly to support various package types. The ICSP™ cable assembly is included as a standard item. In Stand-Alone mode, the MPLAB PM3 Device Programmer can read, verify and program PIC devices without a PC connection. It can also set code protection in this mode. The MPLAB PM3 connects to the host PC via an RS-232 or USB cable. The MPLAB PM3 has high-speed communications and optimized algorithms for quick programming of large memory devices and incorporates an MMC card for file storage and data applications.

27.13 Demonstration/Development Boards, Evaluation Kits, and Starter Kits

A wide variety of demonstration, development and evaluation boards for various PIC MCUs and dsPIC DSCs allows quick application development on fully functional systems. Most boards include prototyping areas for adding custom circuitry and provide application firmware and source code for examination and modification.

The boards support a variety of features, including LEDs, temperature sensors, switches, speakers, RS-232 interfaces, LCD displays, potentiometers and additional EEPROM memory.

The demonstration and development boards can be used in teaching environments, for prototyping custom circuits and for learning about various microcontroller applications.

In addition to the PICDEM[™] and dsPICDEM[™] demonstration/development board series of circuits, Microchip has a line of evaluation kits and demonstration software for analog filter design, KEELOQ[®] security ICs, CAN, IrDA[®], PowerSmart battery management, SEEVAL[®] evaluation system, Sigma-Delta ADC, flow rate sensing, plus many more.

Also available are starter kits that contain everything needed to experience the specified device. This usually includes a single application and debug capability, all on one board.

Check the Microchip web page (www.microchip.com) for the complete list of demonstration, development and evaluation kits.

28.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings^(†)

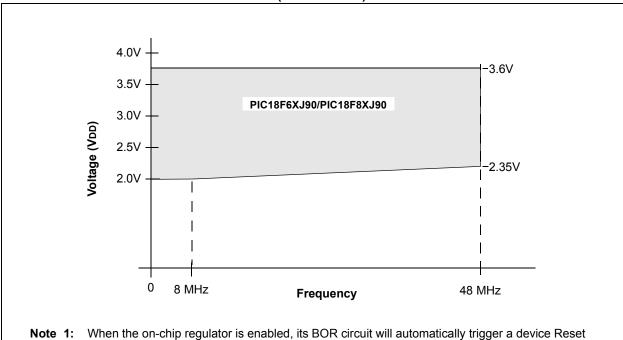
Ambient temperature under bias40°C to +100°C
Storage temperature
Voltage on any digital only I/O pin or MCLR with respect to Vss (except VDD)0.3V to 6.0V
Voltage on any combined digital and analog pin with respect to Vss (except VDD and MCLR)0.3V to (VDD + 0.3V)
Voltage on VDDCORE with respect to Vss0.3V to 2.75V
Voltage on VDD with respect to Vss
Total power dissipation (Note 1)
Maximum current out of Vss pin
Maximum current into VDD pin
Maximum output current sunk by PORTA<7:6> and any PORTB and PORTC I/O pins
Maximum output current sunk by any PORTD, PORTE and PORTJ I/O pins8 mA
Maximum output current sunk by PORTA<5:0> and any PORTF, PORTG and PORTH I/O pins
Maximum output current sourced by PORTA<7:6> and any PORTB and PORTC I/O pins
Maximum output current sourced by any PORTD, PORTE and PORTJ I/O pins8 mA
Maximum output current sourced by PORTA<5:0> and any PORTF, PORTG and PORTH I/O pins
Maximum current sunk by all ports combined

Note 1: Power dissipation is calculated as follows:

Pdis = VDD x {IDD $-\sum$ IOH} + \sum {(VDD - VOH) x IOH} + \sum (VOL x IOL)

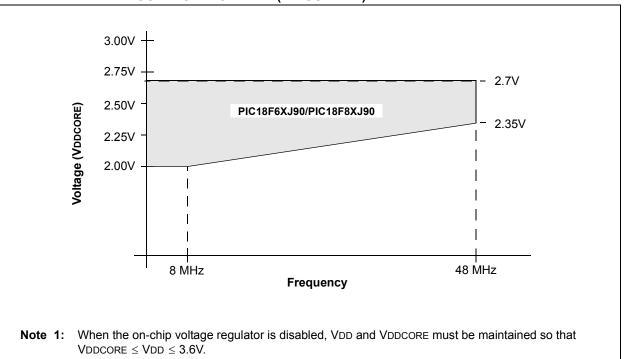
† NOTICE: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

FIGURE 28-1: VOLTAGE-FREQUENCY GRAPH, REGULATOR ENABLED (INDUSTRIAL)⁽¹⁾



before VDD reaches a level at which full-speed operation is not possible.





28.1	DC Characteristics:	Supply Voltage
		PIC18F87J90 Family (Industrial)

PIC18F87J90 Family (Industrial)			$\begin{array}{llllllllllllllllllllllllllllllllllll$				
Param No.	Symbol	Characteristic	Min	Тур	Мах	Units	Conditions
D001	Vdd	Supply Voltage	VDDCORE 2.0		3.6 3.6	V V	ENVREG tied to Vss ENVREG tied to Vpp
D001B	VDDCORE	External Supply for Microcontroller Core	2.0	_	2.70	V	ENVREG tied to Vss
D001C	AVdd	Analog Supply Voltage	VDD - 0.3	_	VDD + 0.3	V	
D001D	AVss	Analog Ground Potential	Vss – 0.3	_	Vss + 0.3	V	
D002	Vdr	RAM Data Retention Voltage ⁽¹⁾	1.5	_	—	V	
D003	Vpor	VDD Start Voltage to Ensure Internal Power-on Reset Signal	—		0.7	V	See Section 5.3 "Power-on Reset (POR)" for details
D004	SVDD	VDD Rise Rate to Ensure Internal Power-on Reset Signal	0.05	_	_	V/ms	See Section 5.3 "Power-on Reset (POR)" for details
D005	VBOR	Brown-out Reset Voltage	—	1.8	_	V	

Note 1: This is the limit to which VDD can be lowered in Sleep mode, or during a device Reset, without losing RAM data.

28.2 DC Characteristics: Power-Down and Supply Current PIC18F87J90 Family (Industrial)

PIC18F87J90 Family (Industrial)		Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial					
Param No.	Device	Тур	Max	Units	Condit	ions	
	Power-Down Current (IPD) ⁽¹⁾						
	All devices	0.4	1.4	μA	-40°C		
		0.1	1.4	μA	+25°C	VDD = 2.0V ⁽⁴⁾	
		0.8	6	μA	+60°C	(Sleep mode)	
		5.5	10.2	μA	+85°C		
	All devices	0.5	1.5	μA	-40°C		
		0.1	1.5	μA	+25°C	VDD = 2.5V ⁽⁴⁾	
		1	8	μA	+60°C	(Sleep mode)	
		6.8	12.6	μA	+85°C		
	All devices	2.9	7	μA	-40°C		
		3.6	7	μA	+25°C	VDD = 3.3V ⁽⁵⁾	
		4.1	10	μA	+60°C	(Sleep mode)	
		9.6	19	μA	+85°C		

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or VSS, and all features that add delta current disabled (such as WDT, Timer1 oscillator, BOR, etc.).

2: The supply current is mainly a function of the operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

- MCLR = VDD; WDT enabled/disabled as specified.
- 3: Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4: Voltage regulator disabled (ENVREG = 0, tied to Vss).
- 5: Voltage regulator enabled (ENVREG = 1, tied to VDD, REGSLP = 1).

	7J90 Family strial)	Standard Operating	-	-	•	otherwise stated) ≨ +85°C for industrial		
Param No.	Device	Тур	Max	Units		Condition	S	
	Supply Current (IDD) ^(2,3)							
	All devices	5	14.2	μA	-40°C			
		5.5	14.2	μA	+25°C	VDD = 2.0V, $VDDCORE = 2.0V(4)$	Fosc = 31 kHz	
		10	19.0	μA	+85°C	VBBOOKE 2.0V		
	All devices	6.8	16.5	μA	-40°C			
		7.6	16.5	μA	+25°C	VDD = 2.5V, VDDCORE = 2.5V ⁽⁴⁾	(RC_RUN mode,	
		14	22.4	μA	+85°C		internal oscillator source	
	All devices	37	84	μA	-40°C			
		51	84	μA	+25°C	VDD = 3.3V ⁽⁵⁾		
		72	108	μA	+85°C			
	All devices	0.43	0.82	mA	-40°C	N/ 0.01/	Fosc = 1 MHz (RC_RUN mode,	
		0.47	0.82	mA	+25°C	VDD = 2.0V, $VDDCORE = 2.0V(4)$		
		0.52	0.95	mA	+85°C			
	All devices	0.52	0.98	mA	-40°C			
		0.57	0.98	mA	+25°C	VDD = 2.5V, $VDDCORE = 2.5V^{(4)}$		
		0.63	1.10	mA	+85°C		internal oscillator source	
	All devices	0.59	0.96	mA	-40°C			
		0.65	0.96	mA	+25°C	VDD = 3.3V ⁽⁵⁾		
		0.72	1.18	mA	+85°C			
	All devices	0.88	1.45	mA	-40°C)/== 0.0)/		
		1	1.45	mA	+25°C	$VDD = 2.0V,$ $VDDCORE = 2.0V^{(4)}$		
		1.1	1.58	mA	+85°C			
	All devices	1.2	1.72	mA	-40°C	(22 - 25)'	Fosc = 4 MHz	
		1.3	1.72	mA	+25°C	VDD = 2.5V, $VDDCORE = 2.5V^{(4)}$	(RC_RUN mode,	
		1.4	1.85	mA	+85°C		internal oscillator source	
	All devices	1.3	2.87	mA	-40°C			
		1.4	2.87	mA	+25°C	VDD = 3.3V ⁽⁵⁾		
		1.5	2.96	mA	+85°C			

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current disabled (such as WDT, Timer1 oscillator, BOR, etc.).

2: The supply current is mainly a function of the operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

3: Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.

4: Voltage regulator disabled (ENVREG = 0, tied to Vss).

5: Voltage regulator enabled (ENVREG = 1, tied to VDD, REGSLP = 1).

28.2 DC Characteristics: Power-Down and Supply Current PIC18F87J90 Family (Industrial) (Continued)

	7J90 Family strial)	Standard Operating				otherwise stated) ≤ +85°C for industrial				
Param No.	Device	Тур	Max	Units		Conditions				
	Supply Current (IDD) Cont.	(2,3)								
	All devices	3	9.4	μA	-40°C					
		3.3	9.4	μA	+25°C	VDD = 2.0V, $VDDCORE = 2.0V(4)$				
		8.5	17.2	μA	+85°C	VBBOOKE 2.0V				
	All devices	4	10.5	μA	-40°C) / 0 5) /	Fosc = 31 kHz			
		4.3	10.5	μA	+25°C	VDD = 2.5V, VDDCORE = 2.5V ⁽⁴⁾	(RC_IDLE mode,			
		10.3	19.5	μA	+85°C	VDDCORE - 2.0V	internal oscillator source			
	All devices	34	82	μA	-40°C					
		48	82	μA	+25°C	VDD = 3.3V ⁽⁵⁾				
		69	105	μA	+85°C					
	All devices	0.33	0.75	mA	-40°C		Fosc = 1 MHz (RC_IDLE mode,			
		0.37	0.75	mA	+25°C	VDD = 2.0V, $VDDCORE = 2.0V^{(4)}$				
		0.41	0.84	mA	+85°C	VDDCORE - 2.0V				
	All devices	0.39	0.78	mA	-40°C) / 0 5) /				
		0.42	0.78	mA	+25°C	VDD = 2.5V, VDDCORE = 2.5V ⁽⁴⁾				
		0.47	0.91	mA	+85°C	VBBOOKE 2.0V	internal oscillator source			
	All devices	0.43	0.82	mA	-40°C					
		0.48	0.82	mA	+25°C	VDD = 3.3V ⁽⁵⁾				
		0.54	0.95	mA	+85°C					
	All devices	0.53	0.98	mA	-40°C)/== 0.0)/				
		0.57	0.98	mA	+25°C	VDD = 2.0V, $VDDCORE = 2.0V(4)$				
		0.61	1.12	mA	+85°C	VBBOOKE 2.0V				
	All devices	0.63	1.14	mA	-40°C	$\lambda (22 - 2.5)$	Fosc = 4 MHz			
		0.67	1.14	mA	+25°C	VDD = 2.5V, VDDCORE = 2.5V ⁽⁴⁾	(RC_IDLE mode,			
		0.72	1.25	mA	+85°C		internal oscillator source)			
	All devices	0.7	1.27	mA	-40°C					
		0.76	1.27	mA	+25°C	VDD = 3.3V ⁽⁵⁾				
		0.82	1.45	mA	+85°C					

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current disabled (such as WDT, Timer1 oscillator, BOR, etc.).

- 2: The supply current is mainly a function of the operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.
 - The test conditions for all IDD measurements in active operation mode are:
 - OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;
 - MCLR = VDD; WDT enabled/disabled as specified.
- **3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4: Voltage regulator disabled (ENVREG = 0, tied to Vss).
- 5: Voltage regulator enabled (ENVREG = 1, tied to VDD, REGSLP = 1).

	7J90 Family strial)	Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial									
Param No.	Device	Тур	Max	Units		Conditions	;				
	Supply Current (IDD) Cont.	(2,3)									
	All devices	0.17	0.35	mA	-40°C						
		0.18	0.35	mA	+25°C	VDD = 2.0V, VDDCORE = 2.0V ⁽⁴⁾					
		0.20	0.42	mA	+85°C	VDDOORE 2.0V					
	All devices	0.29	0.52	mA	-40°C		Fosc = 1 MHz				
		0.31	0.52	mA	+25°C	VDD = 2.5V, VDDCORE = 2.5V ⁽⁴⁾	(PRI_RUN mode, EC oscillator)				
		0.34	0.61	mA	+85°C	VDDOORE 2.0V					
	All devices	0.59	1.1	mA	-40°C						
		0.44	0.85	mA	+25°C	VDD = 3.3V ⁽⁵⁾					
		0.42	0.85	mA	+85°C						
	All devices	0.70	1.25	mA	-40°C						
		0.75	1.25	mA	+25°C	VDD = 2.0V, $VDDCORE = 2.0V(4)$	Fosc = 4 MHz				
		0.79	1.36	mA	+85°C	VDDOORE 2.0V					
	All devices	1.10	1.7	mA	-40°C						
		1.10	1.7	mA	+25°C	VDD = 2.5V, $VDDCORE = 2.5V^{(4)}$	(PRI_RUN mode,				
		1.12	1.82	mA	+85°C		EC oscillator)				
	All devices	1.55	1.95	mA	-40°C						
		1.47	1.89	mA	+25°C	VDD = 3.3V ⁽⁵⁾					
		1.54	1.92	mA	+85°C						
	All devices	9.9	14.8	mA	-40°C	$\lambda = 0.5 \lambda$					
		9.5	14.8	mA	+25°C	VDD = 2.5V, $VDDCORE = 2.5V^{(4)}$					
		10.1	15.2	mA	+85°C		Fosc = 48 MHz (PRI_RUN mode,				
	All devices	13.3	23.2	mA	-40°C		EC oscillator)				
		12.2	22.7	mA	+25°C	VDD = 3.3V ⁽⁵⁾	,				
		12.1	22.7	mA	+85°C						

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current disabled (such as WDT, Timer1 oscillator, BOR, etc.).

2: The supply current is mainly a function of the operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

- **3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4: Voltage regulator disabled (ENVREG = 0, tied to Vss).
- 5: Voltage regulator enabled (ENVREG = 1, tied to VDD, REGSLP = 1).

	7J90 Family strial)	Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial								
Param No.	Device	Тур	Мах	Units		Condition	s			
	Supply Current (IDD) Cont.	(2,3)			·					
	All devices	4.5	5.2	mA	-40°C					
		4.4	5.2	mA	+25°C	VDD = 2.5V, $VDDCORE = 2.5V(4)$				
		4.5	5.2	mA	+85°C	VEDOCIAL 2.0V	Fosc = 4 MHz, 16 MHz internal			
	All devices	5.7	6.7	mA	-40°C	VDD = 3.3V ⁽⁵⁾	(PRI_RUN HSPLL mode)			
		5.5	6.3	mA	+25°C					
		5.3	6.3	mA	+85°C					
	All devices	10.8	13.5	mA	-40°C					
		10.8	13.5	mA	+25°C	VDD = 2.5V, VDDCORE = 2.5V ⁽⁴⁾	Fosc = 10 MHz, 40 MHz internal (PRI_RUN HSPLL mode)			
		9.9	13.0	mA	+85°C					
	All devices	13.4	24.1	mA	-40°C					
		12.3	20.2	mA	+25°C	VDD = 3.3V ⁽⁵⁾				
		11.2	19.5	mA	+85°C					

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current disabled (such as WDT, Timer1 oscillator, BOR, etc.).

2: The supply current is mainly a function of the operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

3: Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.

4: Voltage regulator disabled (ENVREG = 0, tied to Vss).

5: Voltage regulator enabled (ENVREG = 1, tied to VDD, REGSLP = 1).

	7 J90 Family strial)	Standard Operating				otherwise stated) ≤ +85°C for industrial		
Param No.	Device	Тур	Max	Units		Conditions	;	
	Supply Current (IDD) Cont.	(2,3)						
	All devices	0.10	0.26	mA	-40°C			
		0.07	0.18	mA	+25°C	VDD = 2.0V, VDDCORE = 2.0V ⁽⁴⁾		
		0.09	0.22	mA	+85°C	VDDOORE 2.0V		
	All devices	0.25	0.48	mA	-40°C		Fosc = 1 MHz	
		0.13	0.30	mA	+25°C	VDD = 2.5V, VDDCORE = 2.5V ⁽⁴⁾	(PRI_IDLE mode, EC oscillator)	
		0.10	0.26	mA	+85°C	VDDOORE 2.0V		
	All devices	0.45	0.68	mA	-40°C			
		0.26	0.45	mA	+25°C	VDD = 3.3V ⁽⁵⁾		
		0.30	0.54	mA	+85°C			
	All devices	0.36	0.60	mA	-40°C			
		0.33	0.56	mA	+25°C	VDD = 2.0V, $VDDCORE = 2.0V(4)$	Fosc = 4 MHz (PRI_IDLE mode,	
		0.35	0.56	mA	+85°C			
	All devices	0.52	0.81	mA	-40°C			
		0.45	0.70	mA	+25°C	VDD = 2.5V, $VDDCORE = 2.5V^{(4)}$		
		0.46	0.70	mA	+85°C		EC oscillator)	
	All devices	0.80	1.15	mA	-40°C			
		0.66	0.98	mA	+25°C	VDD = 3.3V ⁽⁵⁾		
		0.65	0.98	mA	+85°C			
	All devices	5.2	6.5	mA	-40°C	$\lambda = 0.5 \lambda$		
		4.9	5.9	mA	+25°C	VDD = 2.5V, VDDCORE = 2.5V ⁽⁴⁾		
		3.4	4.5	mA	+85°C		Fosc = 48 MHz (PRI IDLE mode,	
	All devices	6.2	12.4	mA	-40°C		EC oscillator)	
		5.9	11.5	mA	+25°C	VDD = 3.3V ⁽⁵⁾	,	
		5.8	11.5	mA	+85°C			

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current disabled (such as WDT, Timer1 oscillator, BOR, etc.).

2: The supply current is mainly a function of the operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

- **3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4: Voltage regulator disabled (ENVREG = 0, tied to Vss).
- 5: Voltage regulator enabled (ENVREG = 1, tied to VDD, REGSLP = 1).

	7J90 Family strial)	Standard Operating	•	•	•	otherwise stated) +85°C for industrial				
Param No.	Device	Тур	Max	Units		Conditions				
	Supply Current (IDD) Cont.	(2,3)								
	All devices	18	35	μA	-40°C					
		19	35	μA	+25°C	$VDD = 2.0V,$ $VDDCORE = 2.0V^{(4)}$				
		28	49	μA	+85°C	VEBOORE 2.0V	Fosc = 32 kHz ⁽³⁾ (SEC_RUN mode,			
	All devices	20	45	μA	-40°C) (= = = 0.5) (
		21	45	μA	+25°C	VDD = 2.5V, $VDDCORE = 2.5V^{(4)}$				
		32	61	μA	+85°C	VEBOORE 2.0V	Timer1 as clock)			
	All devices	0.06	0.11	mA	-40°C					
		0.07	0.11	mA	+25°C	VDD = 3.3V ⁽⁵⁾				
		0.09	0.15	mA	+85°C					
	All devices	14	28	μA	-40°C					
		15	28	μA	+25°C	$VDD = 2.0V,$ $VDDCORE = 2.0V^{(4)}$				
		24	43	μA	+85°C	VBBOOKE 2.0V				
	All devices	15	31	μA	-40°C		Fosc = 32 kHz ⁽³⁾			
		16	31	μA	+25°C	VDD = 2.5V, $VDDCORE = 2.5V^{(4)}$	(SEC_IDLE mode,			
		27	50	μA	+85°C		Timer1 as clock)			
	All devices	0.05	0.10	mA	-40°C					
		0.06	0.10	mA	+25°C	VDD = 3.3V ⁽⁵⁾				
		0.08	0.14	mA	+85°C					

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current disabled (such as WDT, Timer1 oscillator, BOR, etc.).

2: The supply current is mainly a function of the operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

3: Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.

4: Voltage regulator disabled (ENVREG = 0, tied to Vss).

5: Voltage regulator enabled (ENVREG = 1, tied to VDD, REGSLP = 1).

PIC18F8 (Indus	7J90 Family strial)	Standard (Operating t	•	•	itions (unless ot -40°C ≤ TA ≤ +	herwise stated) ·85°C for industrial		
Param No.	Device	Тур	Max	Units		Condition	s	
D022	Module Differential Curren	ts (Alwdt, A	IOSCB,	Alad)				
	Watchdog Timer	2.1	7.0	μA	-40°C	N== 0.0V/		
		2.2	7.0	μA	+25°C	VDD = 2.0V, $VDDCORE = 2.0V^{(4)}$		
		4.3	9.5	μA	+85°C	VDDCORE - 2.0V		
		3.0	8.0	μA	-40°C			
		3.1	8.0	μA	+25°C	$VDD = 2.5V,$ $VDDCORE = 2.5V^{(4)}$		
		5.5	10.4	μA	+85°C	VDDCORE - 2.3V		
		5.9	12.1	μA	-40°C			
		6.2	12.1	μA	+25°C	VDD = 3.3V		
		6.9	13.6	μA	+85°C			
0024	LCD Module	2 ^(6,7)	5	μA	+25°C	VDD = 2.0V	Resistive Ladder	
∆ILCD)		2.7 ^(6,7)	5	μA	+25°C	VDD = 2.5V	CPEN = 0;	
		3.5 ^(6,7)	7	μA	+25°C	VDD = 3.0V	CKSEL<1:0> = 00; CS<1:0> = 10; LP<3:0> = 0100	
		16 ⁽⁷⁾	25	μA	+25°C	VDD = 2.0V	Charge Pump	
		17 ⁽⁷⁾	25	μA	+25°C	VDD = 2.5V	BIAS<2:0> = 111;	
		24(7)	40	μA	+25°C	VDD = 3.0V	CPEN = 1; CKSEL<1:0> = 11; CS<1:0> = 10	
D025	RTCC + Timer1 Osc. with	0.9	4.0	μA	-10°C	N== 0.0V/		
∆loscb)	32 kHz Crystal ⁽⁶⁾	1.0	4.5	μA	+25°C	$VDD = 2.0V,$ $VDDCORE = 2.0V^{(4)}$	32 kHz on Timer1 ⁽³⁾	
		1.1	4.5	μA	+85°C	VDDCORE = 2.0V		
		1.1	4.5	μA	-10°C			
		1.2	5.0	μA	+25°C	VDD = 2.5V, $VDDCORE = 2.5V^{(4)}$	32 kHz on Timer1 ⁽³⁾	
		1.2	5.0	μA	+85°C			
		1.6	6.5	μA	-10°C			
		1.6	6.5	μA	+25°C	VDD = 3.3V	32 kHz on Timer1 ⁽³⁾	
		2.1	8.0	μA	+85°C			
D026 [ΔΙΑD)	A/D Converter	3.0	10.0	μA	-40°C to +85°C	VDD = 2.0V, $VDDCORE = 2.0V^{(4)}$	A/D on, not converting	
		3.0	10.0	μA	-40°C to +85°C	VDD = 2.5V,		

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or VSS, and all features that add delta current disabled (such as WDT, Timer1 oscillator, BOR, etc.).

2: The supply current is mainly a function of the operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

- **3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4: Voltage regulator disabled (ENVREG = 0, tied to Vss).
- 5: Voltage regulator enabled (ENVREG = 1, tied to VDD, REGSLP = 1).

28.3 DC Characteristics: PIC18F87J90 Family (Industrial)

DC CHA	RACTER	RISTICS				unless otherwise stated) ≤ +85°C for industrial
Param No.	Symbol	Characteristic	Min	Мах	Units	Conditions
	VIL	Input Low Voltage				
		All I/O Ports:				
D030		with TTL Buffer	Vss	0.15 Vdd	V	VDD < 3.3V
D030A			—	0.8	V	$3.3V \leq V\text{DD} \leq 3.6V$
D031		with Schmitt Trigger Buffer	Vss	0.2 VDD	V	
D031A		with RC3 and RC4	Vss	0.3 VDD	V	I ² C™ enabled
D031B			Vss	0.8	V	SMBus enabled
D032		MCLR	Vss	0.2 Vdd	V	
D033		OSC1	Vss	0.3 Vdd	V	HS, HSPLL modes
D033A		OSC1	Vss	0.2 Vdd	V	EC, ECPLL modes
D034		T13CKI	Vss	0.3	V	
	Viн	Input High Voltage				
		I/O Ports with non 5.5V Tolerance:				
D040		with TTL Buffer	0.25 VDD + 0.8V	Vdd	V	VDD < 3.3V
D040A			2.0	Vdd	V	$3.3V \leq V\text{DD} \leq 3.6V$
D041		with Schmitt Trigger Buffer	0.8 Vdd	Vdd	V	
D041A		RC3 and RC4	0.7 Vdd	Vdd	V	I ² C enabled
D041B			2.1	Vdd	V	SMBus enabled
		I/O Ports with 5.5V Tolerance:				
		with TTL Buffer	0.25 VDD + 0.8V	5.5	V	Vdd < 3.3V
			2.0	5.5	V	$3.3V \leq V\text{DD} \leq 3.6V$
		with Schmitt Trigger Buffer	0.8 Vdd	5.5	V	
D042		MCLR	0.8 Vdd	Vdd	V	
D043		OSC1	0.7 Vdd	Vdd	V	HS, HSPLL modes
D043A		OSC1	0.8 Vdd	Vdd	V	EC, ECPLL modes
D044		T13CKI	1.6	Vdd	V	
	lı∟	Input Leakage Current ⁽¹⁾				
D060		I/O Ports with Analog Functions	—	200	nA	$Vss \le VPIN \le VDD,$ Pin at high-impedance
		Digital Only I/O Ports	—	200	nA	$VSS \le VPIN \le 5.5V$
D061		MCLR	—	±1	μA	$Vss \leq V PIN \leq V DD$
D063		OSC1	—	±1	μA	$Vss \leq V \text{PIN} \leq V \text{DD}$
	IPU	Weak Pull-up Current				
D070	IPURB	PORTB Weak Pull-up Current	30	400	μA	Vdd = 3.3V, Vpin = Vss

Note 1: Negative current is defined as current sourced by the pin.

28.3 DC Characteristics: PIC18F87J90 Family (Industrial) (Continued)
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DC CHA	ARACTE	RISTICS				unless otherwise stated) ≤ +85°C for industrial
Param No.	Symbol	Characteristic	Min	Мах	Units	Conditions
	Vol	Output Low Voltage				
D080		I/O Ports:				
		PORTA, PORTF, PORTG, PORTH	-	0.4	V	IOL = 2 mA, VDD = 3.3V, -40°C to +85°C
		PORTD, PORTE, PORTJ	—	0.4	V	IOL = 3.4 mA, VDD = 3.3V, -40°C to +85°C
		PORTB, PORTC	-	0.4	V	IOL = 3.4 mA, VDD = 3.3V, -40°C to +85°C
D083		OSC2/CLKO (EC, ECPLL modes)	-	0.4	V	IOL = 1.6 mA, VDD = 3.3V, -40°C to +85°C
	Vон	Output High Voltage ⁽¹⁾				
D090		I/O Ports:			V	
		PORTA, PORTF, PORTG, PORTH	2.4	—	V	IOн = -2 mA, VDD = 3.3V, -40°C to +85°C
		PORTD, PORTE, PORTJ	2.4	—	V	IOH = -2 mA, VDD = 3.3V, -40°С to +85°С
		PORTB, PORTC	2.4	—	V	IOн = -2 mA, VDD = 3.3V, -40°C to +85°C
D092		OSC2/CLKO (INTOSC, EC, ECPLL modes)	2.4	—	V	IOH = -1 mA, VDD = 3.3V, -40°C to +85°C
		Capacitive Loading Specs on Output Pins				
D100 ⁽⁴⁾	COSC2	OSC2 Pin	_	15	pF	In HS mode when external clock is used to drive OSC1
D101	Сю	All I/O Pins and OSC2	-	50	pF	To meet the AC Timing Specifications
D102	Св	SCL, SDA	_	400	pF	I ² C [™] Specification

Note 1: Negative current is defined as current sourced by the pin.

28.4 DC Characteristics: CTMU Current Source Specifications

DC CH	ARACT	ERISTICS	$\begin{array}{llllllllllllllllllllllllllllllllllll$						
Param No.	Sym	Characteristic	Min	Typ ⁽¹⁾	Мах	Conditions			
	IOUT1	CTMU Current Source, Base Range	—	550	—	nA	CTMUICON<1:0> = 01		
	IOUT2	CTMU Current Source, 10x Range	—	5.5	—	μA	CTMUICON<1:0> = 10		
	IOUT3	CTMU Current Source, 100x Range	—	55	—	μA	CTMUICON<1:0> = 11		

Note 1: Nominal value at center point of current trim range (CTMUICON<7:2> = 000000).

DC CHA	DC CHARACTERISTICS		Standard Operating Conditions (unless otherwise stated) Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial					
Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions	
		Program Flash Memory						
D130	Eр	Cell Endurance	10K	_	—	E/W	-40°C to +85°C	
D131	Vpr	VDD for Read	VMIN	—	3.6	V	VMIN = Minimum operating voltage	
D132B	Vpew	Voltage for Self-Timed Erase or Write operations						
		VDD	2.35		3.6		ENVREG tied to VDD	
		VDDCORE	2.25		2.7	V	ENVREG tied to Vss	
D133A	Tiw	Self-Timed Write Cycle Time	—	2.8	—	ms		
D133B	TIE	Self-Timed Block Erased Cycle Time	-	33	—	ms		
D134	TRETD	Characteristic Retention	20	—	—	Year	Provided no other specifications are violated	
D135	IDDP	Supply Current during Programming	-	3	14	mA		
D140	TWE	Writes per Erase Cycle	—	—	1		For each physical address	

TABLE 28-1: MEMORY PROGRAMMING REQUIREMENTS

† Data in "Typ" column is at 3.3V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

TABLE 28-2: COMPARATOR SPECIFICATIONS

Operating	Operating Conditions: $3.0V \le VDD \le 3.6V$, $-40^{\circ}C \le TA \le +85^{\circ}C$ (unless otherwise stated)											
Param No.	Sym Characteristics		Sym Characteristics		Min	Тур	Мах	Units	Comments			
D300	VIOFF	Input Offset Voltage	—	±5.0	±25	mV						
D301	VICM	Input Common Mode Voltage	0	_	AVDD - 1.5	V						
D302	CMRR	Common Mode Rejection Ratio	55	_	_	dB						
D303	TRESP	Response Time ⁽¹⁾	—	150	400	ns						
D304	Тмс2о∨	Comparator Mode Change to Output Valid*	—	_	10	μS						

Note 1: Response time measured with one comparator input at (AVDD – 1.5)/2, while the other input transitions from Vss to VDD.

TABLE 28-3: VOLTAGE REFERENCE SPECIFICATIONS

Operating Conditions: $3.0V \le V_{DD} \le 3.6V$, -40°C \le TA \le +85°C (unless otherwise stated)										
Param No. Sym Characteristics		Characteristics	Min	Тур	Max	Units	Comments			
D310	VRES	Resolution	VDD/24	_	VDD/32	LSb				
D311	VRAA	Absolute Accuracy	_	_	1/2	LSb				
D312	VRur	Unit Resistor Value (R)	—	2k	—	Ω				
D313	TSET	Settling Time ⁽¹⁾	_	_	10	μS				

Note 1: Settling time measured while CVRR = 1 and CVR<3:0> transitions from '0000' to '1111'.

TABLE 28-4: INTERNAL VOLTAGE REGULATOR SPECIFICATIONS

Operatir	Operating Conditions: $-40^{\circ}C \le TA \le +85^{\circ}C$ (unless otherwise stated)											
Param No.	Sym	Characteristics	Min	Тур	Мах	Units	Comments					
	Vrgout	Regulator Output Voltage	_	2.5	—	V						
	Cefc	External Filter Capacitor Value	4.7	10	_	μF	Capacitor must be low-ESR, a low series resistance (< 5Ω)					

TABLE 28-5: INTERNAL LCD VOLTAGE REGULATOR SPECIFICATIONS

Operatir	Operating Conditions: $2.0V \le VDD \le 3.6V$, $-40^{\circ}C \le TA \le +85^{\circ}C$ (unless otherwise stated)											
Param No.	Sym	Characteristics	Min	Тур	Max	Units	Comments					
	CFLY	Fly Back Capacitor	0.47	4.7		μF	Capacitor must be low-ESR					
	VBIAS			3.40	3.6	V	BIAS<2:0> = 111					
	LCDBIAS3		3.27		V	BIAS<2:0> = 110						
				3.14	—	V	BIAS<2:0> = 101					
				3.01	—	V	BIAS<2:0> = 100					
				2.88	—	V	BIAS<2:0> = 011					
				2.75	—	V	BIAS<2:0> = 010					
				2.62	—	V	BIAS<2:0> = 001					
				2.49		V	BIAS<2:0> = 000					

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28.5 AC (Timing) Characteristics

28.5.1 TIMING PARAMETER SYMBOLOGY

The timing parameter symbols have been created following one of the following formats:

1. TppS2ppS	3	3. Tcc:st	(I ² C specifications only)
2. TppS		4. Ts	(I ² C specifications only)
Т			
F	Frequency	Т	Time
Lowercase le	etters (pp) and their meanings:		
рр			
сс	CCP1	osc	OSC1
ck	CLKO	rd	RD
cs	CS	rw	RD or WR
di	SDI	sc	SCK
do	SDO	ss	SS
dt	Data in	tO	TOCKI
io	I/O port	t1	T13CKI
mc	MCLR	wr	WR
Uppercase le	etters and their meanings:	•	
S			
F	Fall	Р	Period
н	High	R	Rise
I	Invalid (High-impedance)	V	Valid
L	Low	Z	High-impedance
I ² C only			
AA	output access	High	High
BUF	Bus free	Low	Low
TCC:ST (I ² C s	specifications only)		
CC			
HD	Hold	SU	Setup
ST			
DAT	DATA input hold	STO	Stop condition
STA	Start condition		

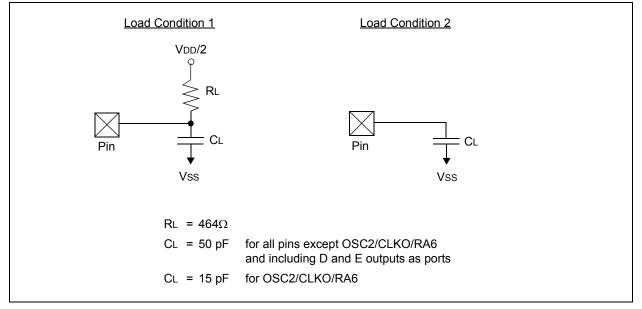
28.5.2 TIMING CONDITIONS

The temperature and voltages specified in Table 28-6 apply to all timing specifications unless otherwise noted. Figure 28-3 specifies the load conditions for the timing specifications.

TABLE 28-6: TEMPERATURE AND VOLTAGE SPECIFICATIONS – AC

	Standard Operating Conditions (unless otherwise stated)
AC CHARACTERISTICS	Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial
	Operating voltage VDD range as described in Section 28.1 and Section 28.3 .

FIGURE 28-3: LOAD CONDITIONS FOR DEVICE TIMING SPECIFICATIONS



28.5.3 TIMING DIAGRAMS AND SPECIFICATIONS

FIGURE 28-4: EXTERNAL CLOCK TIMING

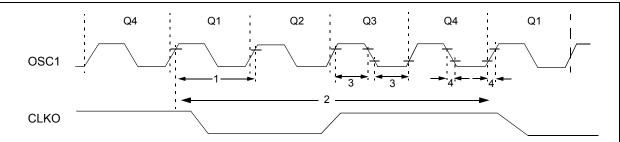


TABLE 28-7:	EXTERNAL CLOCK TIMING REQUIREMENTS

Param. No.	Symbol	Characteristic	Min	Мах	Units	Conditions
1A	Fosc	External CLKI Frequency ⁽¹⁾	DC	48	MHz	EC Oscillator mode
			DC	10		ECPLL Oscillator mode
		Oscillator Frequency ⁽¹⁾	4	25	MHz	HS Oscillator mode
			4	10		HSPLL Oscillator mode
1	Tosc	External CLKI Period ⁽¹⁾	20.8	—	ns	EC Oscillator mode
			100	—		ECPLL Oscillator mode
		Oscillator Period ⁽¹⁾	40.0	250	ns	HS Oscillator mode
			100	250		HSPLL Oscillator mode
2	Тсү	Instruction Cycle Time ⁽¹⁾	83.3	—	ns	Tcy = 4/Fosc, Industrial
3	TosL, TosH	External Clock in (OSC1) High or Low Time	10	—	ns	HS Oscillator mode
4	TosR, TosF	External Clock in (OSC1) Rise or Fall Time	—	7.5	ns	HS Oscillator mode

Note 1: Instruction cycle period (Tcr) equals four times the input oscillator time base period for all configurations except PLL. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min." values with an external clock applied to the OSC1/CLKI pin. When an external clock input is used, the "max." cycle time limit is "DC" (no clock) for all devices.

Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions			
F10	Fosc	Oscillator Frequency Range	4	_	10	MHz	HS mode only			
F11	Fsys	On-Chip VCO System Frequency	16	_	40	MHz	HS mode only			
F12	t _{rc}	PLL Start-up Time (Lock Time)	—	—	2	ms				
F13	∆CLK	CLKO Stability (Jitter)	-2	_	+2	%				

TABLE 28-8:	PLL CLOCK TIMING SPECIFICATIONS (VDD = 2.15V TO 3.6V)
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† Data in "Typ" column is at 3.3V, 25°C, unless otherwise stated. These parameters are for design guidance only and are not tested.

TABLE 28-9: INTERNAL RC ACCURACY (INTOSC AND INTRC SOURCES)

	87J90 Family ustrial)		d Operati g tempera	-		lless otherwise state TA \leq +85°C for industr				
Param No.	Device	Min	Тур	Max	Units	Conditions				
	INTOSC Accuracy @ Freq = 8 MHz, 4 MHz, 2 MHz, 1 MHz, 500 kHz, 250 kHz, 125 kHz, 31 kHz ⁽¹⁾									
	All Devices	-2	+/-1	2	%	+25°C	VDD = 2.7-3.3V			
		-5	_	5	%	-10°C to +85°C	VDD = 2.0-3.3V			
		-10	+/-1	10	%	-40°C to +85°C	VDD = 2.0-3.3V			
	INTRC Accuracy @ Freq	= 31 kHz	(1)							
	All Devices	21.7		40.3	kHz	-40°C to +85°C	VDD = 2.0-3.3V			

Note 1: The accuracy specification of the 31 kHz clock is determined by which source is providing it at a given time. When INTSRC (OSCTUNE<7>) is '1', use the INTOSC accuracy specification. When INTSRC is '0', use the INTRC accuracy specification.

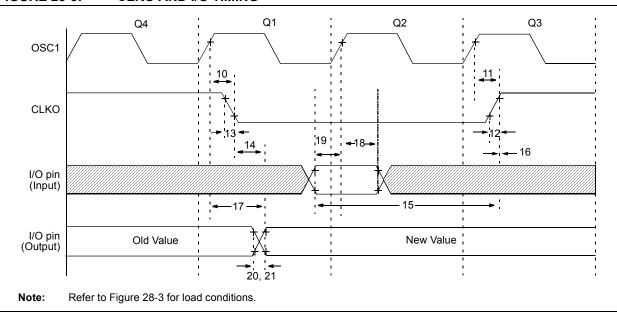


FIGURE 28-5: CLKO AND I/O TIMING

TABLE 28-10: CLKO AND I/O TIMING REQUIREMENTS

Param No.	Symbol	Characteristic	Min	Тур	Max	Units	Conditions
10	TosH2cĸL	OSC1 ↑ to CLKO \downarrow	—	75	200	ns	(Note 1)
11	TosH2ckH	OSC1 ↑ to CLKO ↑	—	75	200	ns	(Note 1)
12	TCKR	CLKO Rise Time	—	15	30	ns	(Note 1)
13	ТскF	CLKO Fall Time	—	15	30	ns	(Note 1)
14	TCKL2IOV	CLKO \downarrow to Port Out Valid	—		0.5 Tcy + 20	ns	
15	ТюV2скН	Port In Valid before CLKO ↑	0.25 Tcy + 25		—	ns	
16	TckH2iol	Port In Hold after CLKO ↑	0		—	ns	
17	TosH2IoV	OSC1 ↑ (Q1 cycle) to Port Out Valid	—	50	150	ns	
18	TosH2ıol	OSC1 ↑ (Q2 cycle) to Port Input Invalid (I/O in hold time)	100	_	—	ns	
19	TioV2osH	Port Input Valid to OSC1 ↑ (I/O in setup time)	0	—	_	ns	
20	TIOR	Port Output Rise Time	—		6	ns	
21	TIOF	Port Output Fall Time	—		5	ns	
22†	TINP	INTx pin High or Low Time	Тсү	_	—	ns	
23†	Trbp	RB<7:4> Change INTx High or Low Time	Тсү		—	ns	

† These parameters are asynchronous events not related to any internal clock edges.

Note 1: Measurements are taken in EC mode, where CLKO output is 4 x Tosc.

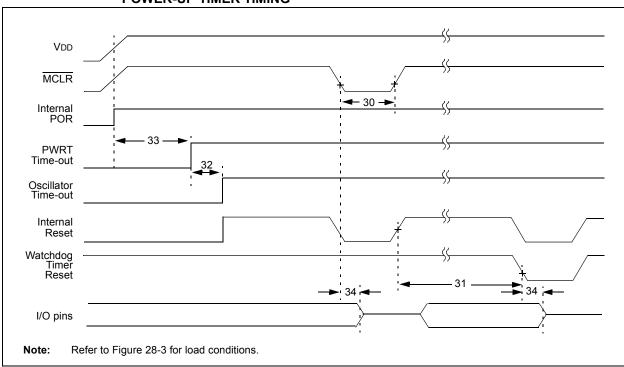


FIGURE 28-6: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND POWER-UP TIMER TIMING

TABLE 28-11: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER, POWER-UP TIMER AND BROWN-OUT RESET REQUIREMENTS

Param. No.	Symbol	Characteristic	Min	Тур	Мах	Units	Conditions
30	ТмсL	MCLR Pulse Width (low)	2 Tcy	10 Тсү	_		(Note 1)
31	Twdt	Watchdog Timer Time-out Period (no postscaler)	3.4	4.0	4.6	ms	
32	Tost	Oscillation Start-up Timer Period	1024 Tosc	_	1024 Tosc		Tosc = OSC1 period
33	TPWRT	Power-up Timer Period	45.8	65.5	85.2	ms	
34	Tioz	I/O High-Impedance from MCLR Low or Watchdog Timer Reset	—	2	—	μs	
38	TCSD	CPU Start-up Time	—	10	_	μs	
				200		μs	Voltage Regulator enabled and put to sleep
39	TIOBST	Time for INTOSC to Stabilize	_	1	—	μs	

Note 1: To ensure device Reset, $\overline{\text{MCLR}}$ must be low for at least 2 TCY or 400 μ s, whichever is lower.

FIGURE 28-7: TIMER0 AND TIMER1 EXTERNAL CLOCK TIMINGS

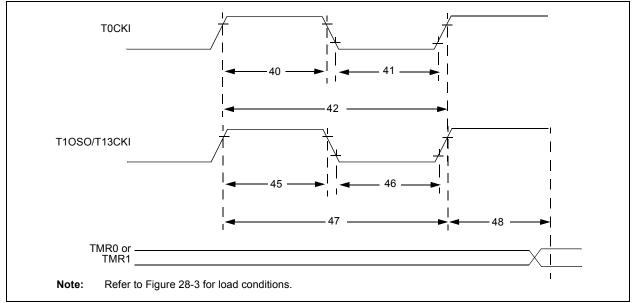


TABLE 28-12:	TIMER0 AND TIMER1 EXTERNAL CLOCK REQUIREMENTS
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Param No.	Symbol		Characteristic		Min	Мах	Units	Conditions
40	T⊤0H	T0CKI High P	ulse Width No prescale		0.5 Tcy + 20	—	ns	
				With prescaler	10	_	ns	
41	T⊤0L	T0CKI Low Pu	ulse Width	No prescaler	0.5 Tcy + 20	_	ns	
				With prescaler	10	_	ns	
42	T⊤0P	T0CKI Period No pre		No prescaler	Tcy + 10	—	ns	
				With prescaler	Greater of: 20 ns or (Tcy + 40)/N	_	ns	N = prescale value (1, 2, 4,, 256)
45	T⊤1H	T13CKI High Synchronous, no prescaler		o prescaler	0.5 Tcy + 20	—	ns	
		Time	Synchronous, with prescaler		10	—	ns	
			Asynchronous		30	_	ns	
46	T⊤1L	T13CKI Low	Synchronous, n	o prescaler	0.5 Tcy + 5	_	ns	
		Time	Synchronous, w	vith prescaler	10	—	ns	
			Asynchronous		30	_	ns	
47	T⊤1P	T13CKI Input Period	Synchronous		Greater of: 20 ns or (Tcy + 40)/N	_	ns	N = prescale value (1, 2, 4, 8)
			Asynchronous		60		ns	
	F⊤1	T13CKI Oscill	ator Input Frequency Range		DC	50	kHz	
48	TCKE2TMRI	Delay from Ex Timer Increme	tternal T13CKI C ent	lock Edge to	2 Tosc	7 Tosc	_	

FIGURE 28-8: CAPTURE/COMPARE/PWM TIMINGS (CCP1, CCP2 MODULES)

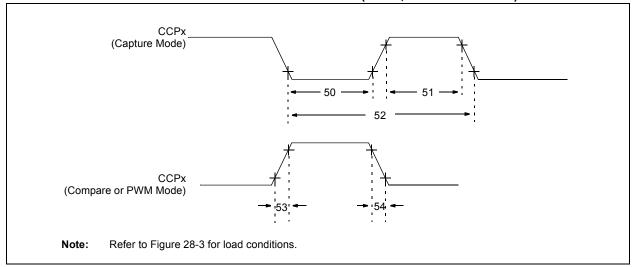


TABLE 28-13: CAPTURE/COMPARE/PWM REQUIREMENTS (CCP1, CCP2 MODULES)

Param No.	Symbol	С	haracteristic	Min	Max	Units	Conditions
50	TccL	CCPx Input Low	No prescaler	0.5 TCY + 20		ns	
		Time	With prescaler	10	_	ns	
51	ТссН	CCPx Input	No prescaler	0.5 Tcy + 20	_	ns	
		High Time	With prescaler	10	_	ns	
52	TCCP	CCPx Input Perio	od	<u>3 Tcy + 40</u> N	_	ns	N = prescale value (1, 4 or 16)
53	TccR	CCPx Output Fal	ll Time	—	25	ns	
54	TccF	CCPx Output Fal	II Time	—	25	ns	

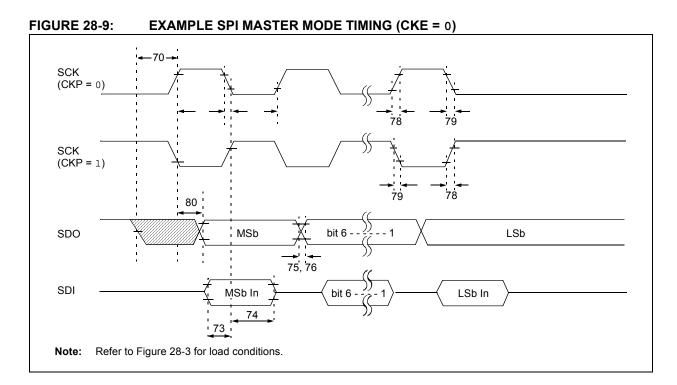


TABLE 28-14: EXAMPLE SPI MODE REQUIREMENTS (MASTER MODE, CKE = 0
--

Param No.	Symbol	Characteristic	Min	Max	Units	Conditions
73	TDIV2scH, TDIV2scL	Setup Time of SDI Data Input to SCK Edge	20	—	ns	
73A	Тв2в	Last Clock Edge of Byte 1 to the 1st Clock Edge of Byte 2	1.5 Tcy + 40		ns	(Note 2)
74	TscH2dlL, TscL2dlL	Hold Time of SDI Data Input to SCK Edge	40	_	ns	
75	TDOR	SDO Data Output Rise Time	—	25	ns	
76	TDOF	SDO Data Output Fall Time	_	25	ns	
78	TscR	SCK Output Rise Time (Master mode)	_	25	ns	
79	TscF	SCK Output Fall Time (Master mode)	—	25	ns	
80	TscH2doV, TscL2doV	SDO Data Output Valid after SCK Edge	—	50	ns	

Note 1: Requires the use of Parameter #73A.

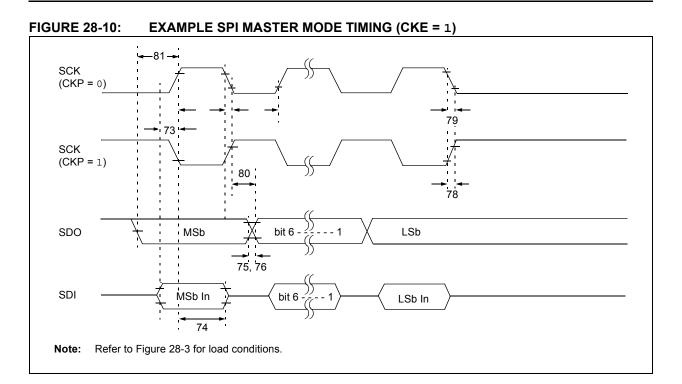
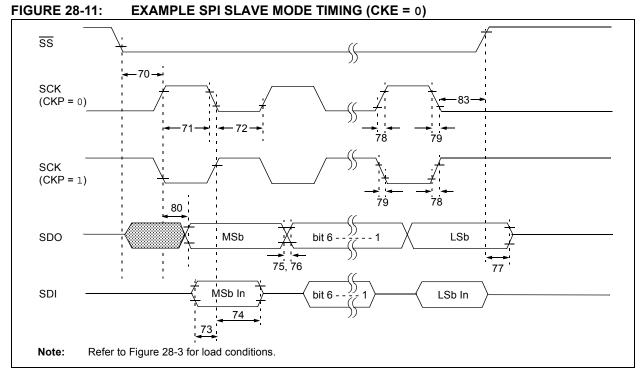


TABLE 28-15: EXAMPLE SPI MODE REQUIREMENTS (MASTER MODE, CKE = 1)

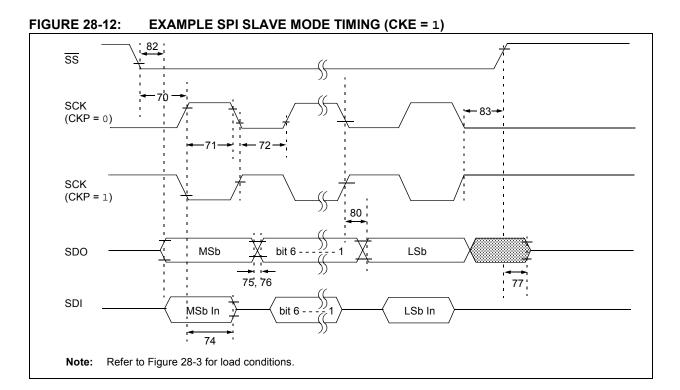
Param. No.	Symbol	Characteristic	Min	Max	Units	Conditions
73	TDIV2scH, TDIV2scL	Setup Time of SDI Data Input to SCK Edge	20	—	ns	
73A	Тв2в	Last Clock Edge of Byte 1 to the 1st Clock Edge of Byte 2	1.5 Tcy + 40	_	ns	(Note 2)
74	TscH2dlL, TscL2dlL	Hold Time of SDI Data Input to SCK Edge	KEdge 40		ns	
75	TDOR	SDO Data Output Rise Time	—	25	ns	
76	TDOF	SDO Data Output Fall Time	—	25	ns	
78	TscR	SCK Output Rise Time (Master mode)	—	25	ns	
79	TscF	SCK Output Fall Time (Master mode)	—	25	ns	
80	TscH2doV, TscL2doV	SDO Data Output Valid after SCK Edge	—	50	ns	
81	TDOV2scH, TDOV2scL	SDO Data Output Setup to SCK Edge	Тсү	—	ns	

Note 1: Requires the use of Parameter #73A.



	20-16: E	XAMPLE SPI MODE REQUIREMEN		, CKE	= = 0)		
Param No.	Symbol	Characteristic		Min	Мах	Units	Conditions
70	TssL2scH, TssL2scL	$\overline{SS} \downarrow$ to SCK \downarrow or SCK \uparrow Input		3 Тсү	—	ns	
70A	TssL2WB	SS to write to SSPBUF		3 TCY	_	ns	
71	TscH	SCK Input High Time	Continuous	1.25 Tcy + 30		ns	
71A		(Slave mode)	Single Byte	40	_	ns	(Note 1)
72	TscL	SCK Input Low Time	Continuous	1.25 Tcy + 30	_	ns	
72A		(Slave mode)	Single Byte	40		ns	(Note 1)
73	TDIV2scH, TDIV2scL	Setup Time of SDI Data Input to SCK Edge		20	_	ns	
73A	Тв2в	Last Clock Edge of Byte 1 to the First Clock Edge of Byte 2		1.5 Tcy + 40		ns	(Note 2)
74	TscH2DIL, TscL2DIL	Hold Time of SDI Data Input to SCK Ed	ge	40	_	ns	
75	TDOR	SDO Data Output Rise Time			25	ns	
76	TDOF	SDO Data Output Fall Time			25	ns	
77	TssH2doZ	SS ↑ to SDO Output High-Impedance		10	50	ns	
78	TscR	SCK Output Rise Time (Master mode)		_	25	ns	
79	TscF	SCK Output Fall Time (Master mode)		_	25	ns	
80	TscH2doV, TscL2doV	SDO Data Output Valid after SCK Edge		_	50	ns	
83	TscH2ssH, TscL2ssH	SS ↑ after SCK Edge		1.5 Tcy + 40	_	ns	

Note 1: Requires the use of Parameter #73A.



Param No.	Symbol	Characteristic		Min	Max	Units	Conditions
70	TssL2scH, TssL2scL	$\overline{SS} \downarrow$ to SCK \downarrow or SCK \uparrow Input		3 Tcy		ns	
70A	TssL2WB	SS to write to SSPBUF		3 Tcy	—	ns	
71	TscH	SCK Input High Time	Continuous	1.25 Tcy + 30		ns	
71A		(Slave mode)	Single Byte	40		ns	(Note 1)
72	TscL	SCK Input Low Time	Continuous	1.25 Tcy + 30	_	ns	
72A		(Slave mode)	Single Byte	40	—	ns	(Note 1)
73A	Тв2в	Last Clock Edge of Byte 1 to the First	1.5 Tcy + 40	_	ns	(Note 2)	
74	TscH2DIL, TscL2DIL	Hold Time of SDI Data Input to SCK Edge		40	—	ns	
75	TDOR	SDO Data Output Rise Time		—	25	ns	
76	TDOF	SDO Data Output Fall Time		—	25	ns	
77	TssH2doZ	SS ↑ to SDO Output High-Impedan	e	10	50	ns	
78	TscR	SCK Output Rise Time (Master mod	e)	—	25	ns	
79	TscF	SCK Output Fall Time (Master mode)		—	25	ns	
80	TscH2doV, TscL2doV	SDO Data Output Valid after SCK Edge		—	50	ns	
82	TssL2DoV	SDO Data Output Valid after $\overline{\text{SS}} \downarrow \text{Edge}$		_	50	ns	
83	TscH2ssH, TscL2ssH	SS ↑ after SCK Edge		1.5 Tcy + 40	_	ns	

Note 1: Requires the use of Parameter #73A.

FIGURE 28-13: I²C[™] BUS START/STOP BITS TIMING

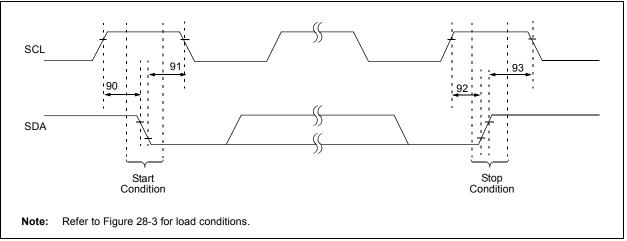
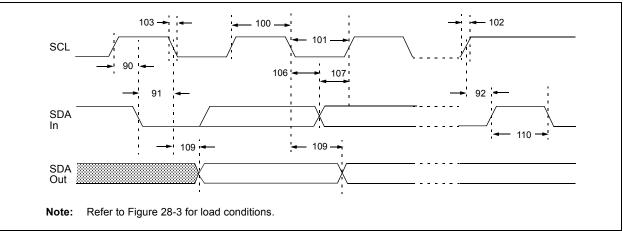


TABLE 28-18: I²C[™] BUS START/STOP BITS REQUIREMENTS (SLAVE MODE)

Param. No.	Symbol	Characte	ristic	Min	Мах	Units	Conditions
90	TSU:STA	Start Condition	100 kHz mode	4700	_	ns	Only relevant for Repeated
		Setup Time	400 kHz mode	600	_		Start condition
91	THD:STA	Start Condition	100 kHz mode	4000	_	ns	After this period, the first
		Hold Time	400 kHz mode	600	_		clock pulse is generated
92	Tsu:sto	Stop Condition	100 kHz mode	4700	—	ns	
		Setup Time	400 kHz mode	600	_		
93	THD:STO	Stop Condition	100 kHz mode	4000	_	ns	
		Hold Time	400 kHz mode	600	_		





Param. No.	Symbol	Characteris	tic	Min	Max	Units	Conditions
100	Тнідн	Clock High Time	100 kHz mode	4.0		μs	
			400 kHz mode	0.6	_	μs	
			MSSP Module	1.5 TCY	—		
101	TLOW	Clock Low Time	100 kHz mode	4.7	—	μs	
			400 kHz mode	1.3	—	μs	
			MSSP Module	1.5 TCY	—		
102	TR	SDA and SCL Rise Time	100 kHz mode	—	1000	ns	
			400 kHz mode	20 + 0.1 CB	300	ns	CB is specified to be from 10 to 400 pF
103	TF	SDA and SCL Fall Time	100 kHz mode	—	300	ns	
			400 kHz mode	20 + 0.1 CB	300	ns	CB is specified to be from 10 to 400 pF
90	TSU:STA	Start Condition Setup Time	100 kHz mode	4.7		μs	Only relevant for Repeated
			400 kHz mode	0.6	—	μs	Start condition
91	THD:STA	Start Condition Hold Time	100 kHz mode	4.0	—	μs	After this period, the first clock
			400 kHz mode	0.6	—	μs	pulse is generated
106	THD:DAT	Data Input Hold Time	100 kHz mode	0	_	ns	
			400 kHz mode	0	0.9	μS	
107	TSU:DAT	Data Input Setup Time	100 kHz mode	250	—	ns	(Note 2)
			400 kHz mode	100	—	ns	
92	Tsu:sto	Stop Condition Setup Time	100 kHz mode	4.7	_	μS	
			400 kHz mode	0.6	_	μS	
109	ΤΑΑ	Output Valid from Clock	100 kHz mode	—	3500	ns	(Note 1)
			400 kHz mode			ns	
110	TBUF	Bus Free Time	100 kHz mode	4.7	—	μS	Time the bus must be free before
			400 kHz mode	1.3	_	μS	a new transmission can start
D102	Св	Bus Capacitive Loading			400	pF	

TABLE 28-19: I ² C [™] BUS DATA REQUIREMENTS (SLAVE MC	DE)
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Note 1: As a transmitter, the device must provide this internal minimum delay time to bridge the undefined region (min. 300 ns) of the falling edge of SCL to avoid unintended generation of Start or Stop conditions.

2: A Fast mode I²C[™] bus device can be used in a Standard mode I²C bus system, but the requirement, TSU:DAT ≥ 250 ns, must then be met. This will automatically be the case if the device does not stretch the LOW period of the SCL signal. If such a device does stretch the LOW period of the SCL signal, it must output the next data bit to the SDA line, TR max. + TSU:DAT = 1000 + 250 = 1250 ns (according to the Standard mode I²C bus specification), before the SCL line is released.

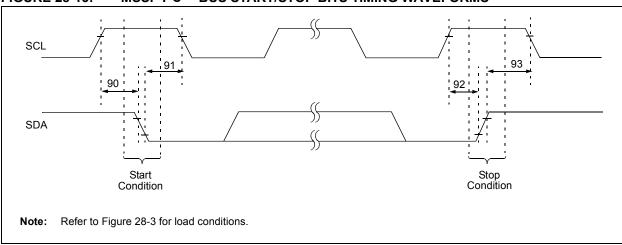
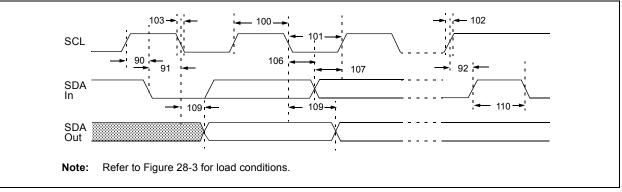


FIGURE 28-15: MSSP I²C[™] BUS START/STOP BITS TIMING WAVEFORMS

Param. No.	Symbol	Characteristic		Min	Max	Units	Conditions
90	TSU:STA	Start Condition	100 kHz mode	2(Tosc)(BRG + 1)		ns	Only relevant for
		Setup Time	400 kHz mode	2(Tosc)(BRG + 1)	_		Repeated Start
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)	_		condition
91	THD:STA	Start Condition	100 kHz mode	2(Tosc)(BRG + 1)	_	ns	After this period, the
		Hold Time	400 kHz mode	2(Tosc)(BRG + 1)	_		first clock pulse is
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)	_		generated
92	Tsu:sto	Stop Condition	100 kHz mode	2(Tosc)(BRG + 1)	_	ns	
		Setup Time	400 kHz mode	2(Tosc)(BRG + 1)	_		
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)	_		
93	THD:STO	Stop Condition	100 kHz mode	2(Tosc)(BRG + 1)	_	ns	
		Hold Time	400 kHz mode	2(Tosc)(BRG + 1)	_	1	
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)	_	1	

Note 1: Maximum pin capacitance = 10 pF for all I^2C^{TM} pins.





Param. No.	Symbol	Charac	teristic	Min	Max	Units	Conditions
100			100 kHz mode	2(Tosc)(BRG + 1)	_	ms	
		Time	400 kHz mode	2(Tosc)(BRG + 1)	_	ms	
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)	_	ms	
101	TLOW	Clock Low Time	100 kHz mode	2(Tosc)(BRG + 1)		ms	
			400 kHz mode	2(Tosc)(BRG + 1)		ms	
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)		ms	
102	Tr	SDA and SCL	100 kHz mode	_	1000	ns	CB is specified to be from
		Rise Time	400 kHz mode	20 + 0.1 Св	300	ns	10 to 400 pF
			1 MHz mode ⁽¹⁾	_	300	ns	
103	TF	SDA and SCL	100 kHz mode		300	ns	CB is specified to be from
		Fall Time	400 kHz mode	20 + 0.1 Св	300	ns	10 to 400 pF
			1 MHz mode ⁽¹⁾	_	100	ns	
90	TSU:STA	Start Condition Setup Time	100 kHz mode	2(Tosc)(BRG + 1)	_	ms	Only relevant for Repeated
			400 kHz mode	2(Tosc)(BRG + 1)	_	ms	Start condition
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)	_	ms	
91	THD:STA	Start Condition	100 kHz mode	2(Tosc)(BRG + 1)		ms	After this period, the first
		Hold Time	400 kHz mode	2(Tosc)(BRG + 1)	_	ms	clock pulse is generated
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)		ms	
106	THD:DAT	Data Input	100 kHz mode	0		ns	
		Hold Time	400 kHz mode	0	0.9	ms	
107	TSU:DAT	Data Input	100 kHz mode	250		ns	(Note 2)
		Setup Time	400 kHz mode	100		ns	
92	TSU:STO	Stop Condition	100 kHz mode	2(Tosc)(BRG + 1)		ms	
		Setup Time	400 kHz mode	2(Tosc)(BRG + 1)		ms	
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)		ms	
109	ΤΑΑ	Output Valid	100 kHz mode	—	3500	ns	
		from Clock	400 kHz mode	—	1000	ns	
			1 MHz mode ⁽¹⁾	—	—	ns	
110	TBUF	Bus Free Time	100 kHz mode	4.7		ms	Time the bus must be free
			400 kHz mode	1.3	_	ms	before a new transmission can start
D102	Св	Bus Capacitive L	oading	_	400	pF	

TABLE 28-21:	MSSP I ² C™	BUS DATA	REQUIREMENTS
--------------	------------------------	-----------------	--------------

Note 1: Maximum pin capacitance = 10 pF for all I^2C^{TM} pins.

2: A Fast mode I²C bus device can be used in a Standard mode I²C bus system, but parameter #107 ≥ 250 ns must then be met. This will automatically be the case if the device does not stretch the LOW period of the SCL signal. If such a device does stretch the LOW period of the SCL signal, it must output the next data bit to the SDA line, parameter #102 + parameter #107 = 1000 + 250 = 1250 ns (for 100 kHz mode), before the SCL line is released.

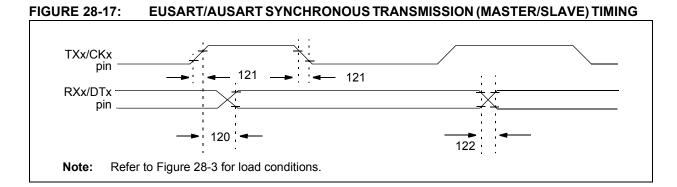


TABLE 28-22: EUSART/AUSART SYNCHRONOUS TRANSMISSION REQUIREMENTS

Param No.	Symbol	Characteristic	Min	Max	Units	Conditions
120	TCKH2DTV	<u>SYNC XMIT (MASTER and SLAVE)</u> Clock High to Data Out Valid		40	ns	
121	TCKRF	Clock Out Rise Time and Fall Time (Master mode)	—	20	ns	
122	TDTRF	Data Out Rise Time and Fall Time		20	ns	

FIGURE 28-18: EUSART/AUSART SYNCHRONOUS RECEIVE (MASTER/SLAVE) TIMING

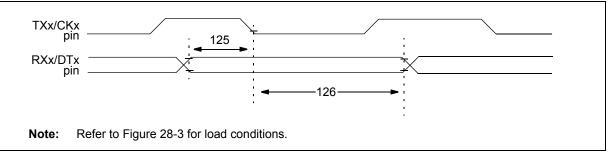


TABLE 28-23: EUSART/AUSART SYNCHRONOUS RECEIVE REQUIREMENTS

Param. No.	Symbol	Characteristic	Min	Max	Units	Conditions
125	TDTV2CKL	SYNC RCV (MASTER and SLAVE)				
		Data Hold before CKx \downarrow (DTx hold time)	10	—	ns	
126	TCKL2DTL	Data Hold after CKx \downarrow (DTx hold time)	15	_	ns	

Param No.	Symbol	Characteristic	Min	Тур	Max	Units	Conditions		
A01	NR	Resolution	_		10	bits			
A03	EIL	Integral Linearity Error	_	_	<±1	LSb	$\Delta VREF \ge 3.0V$		
A04	Edl	Differential Linearity Error	—	_	<±1	LSb	$\Delta \text{VREF} \geq 3.0 \text{V}$		
A06	EOFF	Offset Error	—	_	<±3	LSb	$\Delta VREF \ge 3.0V$		
A07	Egn	Gain Error	—	_	<±3	LSb	$\Delta VREF \ge 3.0V$		
A10	_	Monotonicity	Gi	Guaranteed		aranteed ⁽¹⁾			$VSS \leq VAIN \leq VREF$
A20	$\Delta VREF$	Reference Voltage Range (VREFH – VREFL)	2.0 2.0	_		V V	VDD < 3.0V VDD > 3.0V		
A21	Vrefh	Reference Voltage High	VSS + Δ VREF	_	Vdd	V			
A22	Vrefl	Reference Voltage Low	Vss – 0.3V	_	VDD - 3.0V	V			
A25	VAIN	Analog Input Voltage	VREFL		VREFH	V			
A30	ZAIN	Recommended Impedance of Analog Voltage Source	—		2.5	kΩ			
A50	IREF	VREF Input Current ⁽²⁾		_	5 150	μΑ μΑ	During VAIN acquisition. During A/D conversion cycle.		

 TABLE 28-24:
 A/D CONVERTER CHARACTERISTICS:
 PIC18F87J90 FAMILY (INDUSTRIAL)

Note 1: The A/D conversion result never decreases with an increase in the input voltage and has no missing codes.

2: VREFH current is from RA3/AN3/VREF+ pin or VDD, whichever is selected as the VREFH source. VREFL current is from RA2/AN2/VREF- pin or VSS, whichever is selected as the VREFL source.

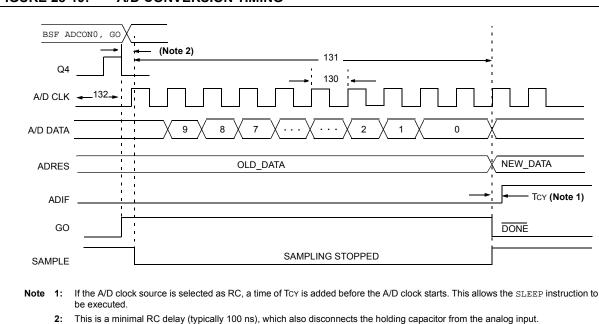


FIGURE 28-19: A/D CONVERSION TIMING

TABLE 28-25: A/D CONVERSION REQUIREMENTS

Param No.	Symbol	Characteristic	Min	Max	Units	Conditions
130	Tad	A/D Clock Period	0.7	25.0 ⁽¹⁾	μS	Tosc based, VREF \geq 3.0V
			—	1	μS	A/D RC mode
131	TCNV	Conversion Time (not including acquisition time) ⁽²⁾	11	12	Tad	
132	TACQ	Acquisition Time ⁽³⁾	1.4		μS	-40°C to +85°C
135	Tswc	Switching Time from Convert \rightarrow Sample	—	(Note 4)		
136	TDIS	Discharge Time	0.2	_	μS	

Note 1: The time of the A/D clock period is dependent on the device frequency and the TAD clock divider.

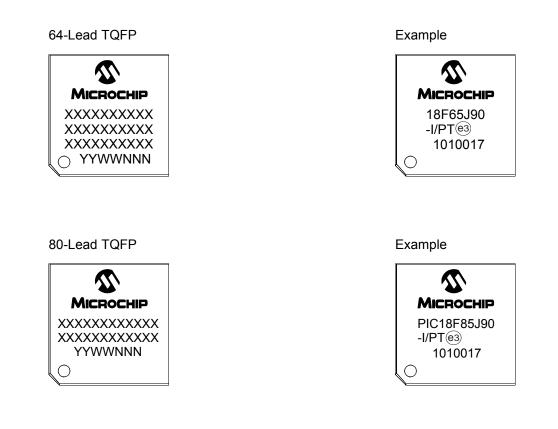
2: ADRES registers may be read on the following TCY cycle.

3: The time for the holding capacitor to acquire the "New" input voltage when the voltage changes full scale after the conversion (VDD to Vss or Vss to VDD). The source impedance (Rs) on the input channels is 50 Ω .

4: On the following cycle of the device clock.

29.0 PACKAGING INFORMATION

29.1 Package Marking Information



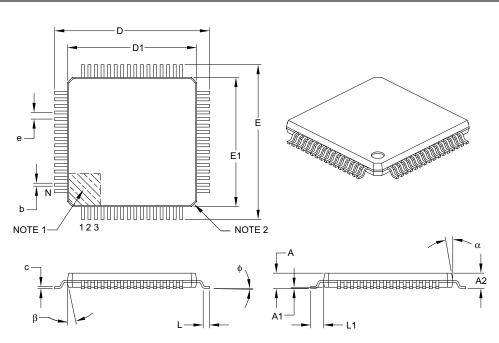
Legend	XXX Y YY WW NNN @3 *	Customer-specific information Year code (last digit of calendar year) Year code (last 2 digits of calendar year) Week code (week of January 1 is week '01') Alphanumeric traceability code Pb-free JEDEC designator for Matte Tin (Sn) This package is Pb-free. The Pb-free JEDEC designator (e3) can be found on the outer packaging for this package.
	be carrie	nt the full Microchip part number cannot be marked on one line, it will d over to the next line, thus limiting the number of available s for customer-specific information.

29.2 Package Details

The following sections give the technical details of the packages.

64-Lead Plastic Thin Quad Flatpack (PT) – 10x10x1 mm Body, 2.00 mm [TQFP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units	MILLIMETERS			
	MIN	NOM	MAX		
Number of Leads	N		64		
Lead Pitch	е		0.50 BSC		
Overall Height	А	_	_	1.20	
Molded Package Thickness	A2	0.95	1.00	1.05	
Standoff	A1	0.05	—	0.15	
Foot Length	L	0.45	0.60	0.75	
Footprint	L1	1.00 REF			
Foot Angle	φ	0°	3.5°	7°	
Overall Width	E		12.00 BSC		
Overall Length	D		12.00 BSC		
Molded Package Width	E1		10.00 BSC		
Molded Package Length	D1		10.00 BSC		
Lead Thickness	С	0.09	_	0.20	
Lead Width	b	0.17	0.22	0.27	
Mold Draft Angle Top	α	11° 12° 13°			
Mold Draft Angle Bottom	β	11°	12°	13°	

Notes:

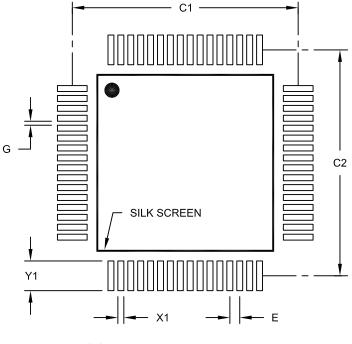
- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. Chamfers at corners are optional; size may vary.
- 3. Dimensions D1 and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.25 mm per side.
- 4. Dimensioning and tolerancing per ASME Y14.5M.
 - BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-085B

64-Lead Plastic Thin Quad Flatpack (PT) – 10x10x1 mm Body, 2.00 mm [TQFP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

	MILLIM	ETERS		
Dimension	MIN	NOM	MAX	
Contact Pitch		0.50 BSC		
Contact Pad Spacing	C1		11.40	
Contact Pad Spacing	C2		11.40	
Contact Pad Width (X64)	X1			0.30
Contact Pad Length (X64)	Y1			1.50
Distance Between Pads	G	0.20		

Notes:

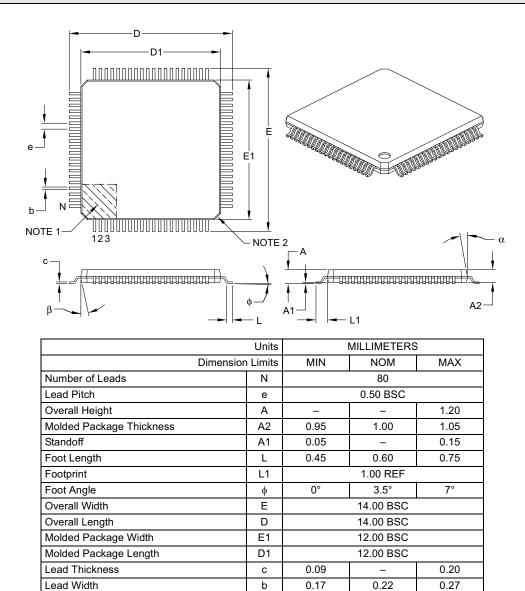
1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2085A

80-Lead Plastic Thin Quad Flatpack (PT) – 12x12x1 mm Body, 2.00 mm [TQFP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

2. Chamfers at corners are optional; size may vary.

Mold Draft Angle Top

Mold Draft Angle Bottom

3. Dimensions D1 and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.25 mm per side.

α

β

11

11°

12°

12°

- 4. Dimensioning and tolerancing per ASME Y14.5M.
 - BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

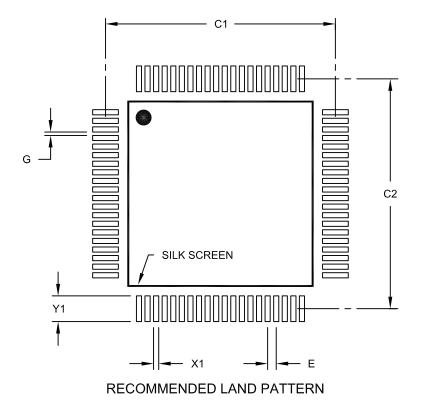
Microchip Technology Drawing C04-092B

13°

13°

80-Lead Plastic Thin Quad Flatpack (PT) – 12x12x1 mm Body, 2.00 mm [TQFP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E	0.50 BSC		
Contact Pad Spacing	C1		13.40	
Contact Pad Spacing	C2		13.40	
Contact Pad Width (X80)	X1			0.30
Contact Pad Length (X80)	Y1			1.50
Distance Between Pads	G	0.20		

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2092A

NOTES:

APPENDIX A: REVISION HISTORY

Revision A (October 2008)

Original data sheet for PIC18F87J90 family devices.

Revision B (December 2008)

Changes to Section 28.2 "DC Characteristics: Power-Down and Supply Current PIC18F87J90 Family (Industrial)" and removal of Section 27.5 "DC Characteristics: RTCC Power-Down Current (IPD)".

Revision C (February 2009)

Added register CONFIG3L to **Section 25.0 "Special Features of the CPU"** and made minor corrections.

Revision D (January 2010)

Added 60°C IPD specification to **Section 28.0 "Electrical Characteristics"**. Removed Preliminary condition tag. Minor edits to text throughout the document.

APPENDIX B: MIGRATION FROM PIC18F85J90 TO PIC18F87J90

Devices in the PIC18F87J90 and PIC18F85J90 families are almost similar in their functions and features. Code can be migrated from the 18F85J90 to the PIC18F87J90 without many changes. The differences between the two device families are listed in Table B-1.

Characteristic	PIC18F87J90 Family	18F85J90 Family
Max Operating Frequency	48 MHz	40 MHz
Max Program Memory	128 Kbytes	32 Kbytes
Data Memory	3,923 Bytes	2,048 Bytes
Program Memory Endurance	10,000 Write/Erase (minimum)	1,000 Write/Erase (minimum)
Single-Word Write for Flash	Yes	No
Oscillator Options	PLL can be used with INTOSC	PLL cannot be used with INTOSC
СТМИ	Yes	No
RTCC	Yes	No
Timer1 Oscillator Options	Low-power oscillator option for Timer1	No
TICKI Clock	T1CKI can be used as a clock without enabling the Timer1 oscillator	No

TABLE B-1: NOTABLE DIFFERENCES BETWEEN PIC18F87J90 AND PIC18F85J90 FAMILIES

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Device ^(1,2)	PIC18F66J90, PIC18F66J90T PIC18F67J90, PIC18F67J90T PIC18F86J90, PIC18F86J90T PIC18F87J90, PIC18F87J90T	
Temperature Range	I = -40°C to +85°C (Industrial)	
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