

# 120mΩ, 1.3A Power Switch with Programmable Current Limit

### **General Description**

The RT9728A is a cost effective, low voltage, single P-MOSFET high side power switch IC for USB application with a programmable current limit feature. Low switch-on resistance (typ.120m $\Omega$ ) and low supply current (typ. 120µA) are realized in this IC. The RT9728A can offer a programmable current limit threshold between 75mA and 1.3A (typ.) via an external resistor. The  $\pm 10\%$  current limit accuracy can be realized for all current limit settings. In addition, a flag output is available to indicate fault conditions to the local USB controller. Furthermore, the chip also integrates an embedded delay function to prevent mis-operation from happening due to high inrush current. The RT9728A is an ideal solution for USB power supply and can support flexible applications since it is functional for various current limit requirements. It is available in SOT-23-6 and WDFN-6L 2x2 packages.

## **Ordering Information**

RT9728A Package Type
E: SOT-23-6
QW: WDFN-6L 2x2
Lead Plating System
G: Green (Halogen Free and Pb Free)
Z: ECO (Ecological Element with
Halogen Free and Pb free)
H: Chip Enable High
L: Chip Enable Low

#### Note:

Richtek products are:

- RoHS compliant and compatible with the current requirements of IPC/JEDEC J-STD-020.
- ▶ Suitable for use in SnPb or Pb-free soldering processes.

### **Features**

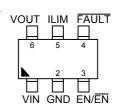
- ±10% Current Limit Accuracy @ 1.3A
- Adjustable Current Limit: 75mA to 1.3A (typ.)
- Meets USB Current Limiting Requirements
- Operating Voltage Range: 2.5V to 5.5V
- Reverse Input-Output Voltage Protection
- Built-in Soft-Start
- 120mΩ High Side MOSFET
- 120μA Supply Current
- RoHS Compliant and Halogen Free

### **Applications**

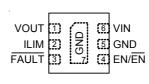
- USB Bus/Self Powered Hubs
- USB Peripheral Ports
- ACPI Power Distribution
- Battery Power Equipment
- 3G/3.5G Data Card, Set-Top Boxes

## **Pin Configurations**

(TOP VIEW)



SOT-23-6



WDFN-6L 2x2



# **Marking Information**

RT9728AHGE

01=DNN

01=: Product Code DNN: Date Code

RT9728ALGE

02=DNN

02=: Product Code DNN: Date Code

RT9728AHGQW



17: Product Code W: Date Code

RT9728ALGQW



19: Product Code W: Date Code

RT9728AHZQW



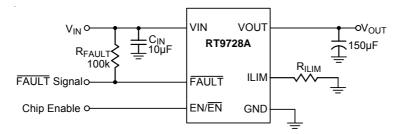
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RT9728ALZQW



19: Product Code W : Date Code

## **Typical Application Circuit**



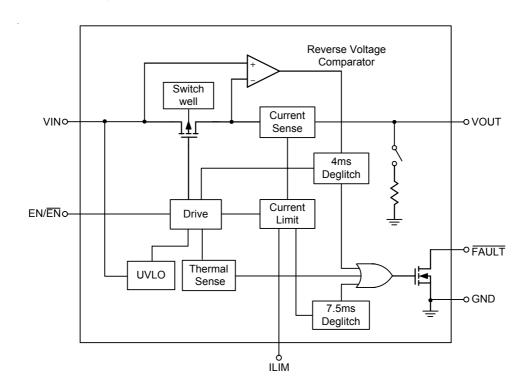
## **Functional Pin Description**

Pin No.		Pin Name	Pin Function		
SOT-23-6	WDFN-6L 2x2	Pin Name	Fill Function		
1	6	VIN	Input Voltage.		
2	5, 7 (Exposed Pad)	GND	Ground. The exposed pad must be soldered to a large PCB and connected to GND for maximum power dissipation.		
3	4	EN/EN	Chip Enable.		
4	3	FAULT	Active-Low Open-Drain Output. Asserted during over current, over temperature, or reverse-voltage conditions.		
5	2	ILIM	Current Limit Set Pin. External resistor used to set current limit threshold. Recommend $19.1k\Omega \le R_{ILIM} \le 232k\Omega$ .		
6	1	VOUT	Power Switch Output.		

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# **Function Block Diagram**





## Absolute Maximum Ratings (Note 1)

• Supply Input Voltage • Other Pin Voltage	
• Power Dissipation, P <sub>D</sub> @ T <sub>A</sub> = 25°C	
SOT-23-6	
• Package Thermal Resistance (Note 2) SOT-23-6, θ <sub>JA</sub>	250°€\\
WDFN-6L 2x2, $\theta_{JA}$	
WDFN-6L 2x2, $\theta_{JC}$	
Junction Temperature	150°C
<ul><li>Storage Temperature Range</li><li>ESD Susceptibility (Note 3)</li></ul>	–65°C to 150°C
HBM (Human Body Model)	2kV
Recommended Operating Conditions (Note 4)	
Supply Input Voltage, VIN	2.5V to 5.5V

### **Electrical Characteristics**

(V<sub>IN</sub> = 3.6V, 19.1k $\Omega$  $\le$  R<sub>ILIM</sub> $\le$  232k $\Omega$ , T<sub>A</sub> = 25°C, unless otherwise specified)

Parameter		Symbol	Test Conditions		Min	Тур	Max	Unit
EN Land Mallana	Logic-High	V <sub>IH</sub>			1.1			V
EN Input Voltage	Logic- Low	VIL					0.66	
Current Limit Threshold Resistor Range		R <sub>ILIM</sub>	(nominal 1%) from I <sub>LIM</sub> to GND		19.1	-	232	kΩ
Under Voltage Lock	cout	V. n. a. o	V <sub>IN</sub> Rising		I	2.3	I	V
Threshold		Vuvlo	V <sub>IN</sub> Falling		I	2.1	I	
Shutdown Current		I <sub>SHDN</sub>	V <sub>IN</sub> = 5.5V, No Load on V <sub>OUT</sub> , V <sub>EN</sub> = 0V			1	3	μА
Quiescent Current		IQ	V <sub>IN</sub> = 5.5V, No Load on V <sub>OUT</sub>	$R_{ILIM} = 20k\Omega$		120	170	μА
				$R_{ILIM} = 210k\Omega$		120	170	
Reverse Leakage Current		I <sub>REV</sub>	V <sub>OUT</sub> = 5.5V, V <sub>IN</sub> = 0V			1	3	μΑ
Thermal Shutdown Temperature		T <sub>SD</sub>			1	160	1	°C
Static Drain-Source On-State Resistance		R <sub>DS(ON)</sub>	I <sub>SW</sub> = 0.2A		1	120	1	mΩ
Current Limit			$R_{ILIM} = 20k\Omega$		1166	1295	1425	
		1	$R_{ILIM} = 49.9k\Omega$		468	520	572	mA
		I <sub>LIM</sub>	R <sub>ILIM</sub> = 210kΩ		110	130	150	
			ILIM Shorted to VIN		50	75	100	

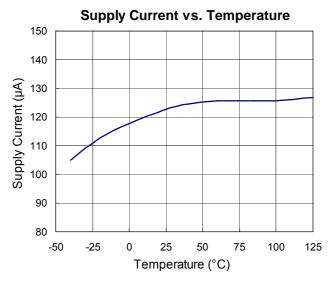


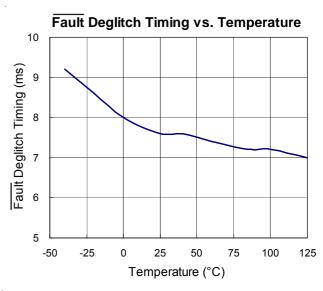
Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Reverse Voltage Comparator Trip Point (V <sub>OUT</sub> – V <sub>IN</sub> )				135		mV
FAULT Output Low Voltage	V <sub>OL</sub>	IFAULT = 1mA	I	180	1	mV
FAULT Off State Leakage		V <sub>FAULT</sub> = 5.5V		1		μΑ
FAULT Deglitch		FAULT assertion or de-assertion due to over current condition	5	7.5	10	mo
Trice Bogillon		FAULT assertion or de-assertion due to reverse voltage condition	2	4	6	ms

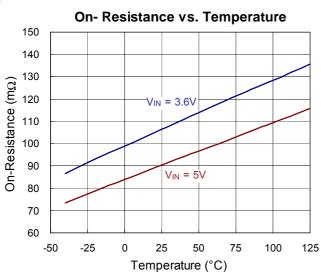
- **Note 1.** Stresses beyond those listed "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.
- Note 2.  $\theta_{JA}$  is measured at  $T_A = 25^{\circ}C$  on a low effective thermal conductivity single-layer test board per JEDEC 51-3.  $\theta_{JC}$  is measured at the exposed pad of the package.
- Note 3. Devices are ESD sensitive. Handling precaution is recommended.
- Note 4. The device is not guaranteed to function outside its operating conditions.

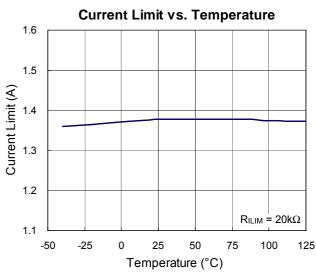


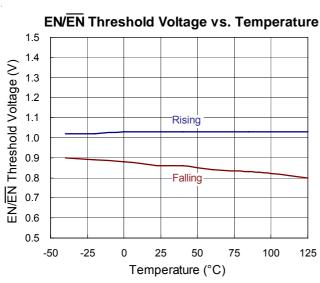
# **Typical Operating Characteristics**

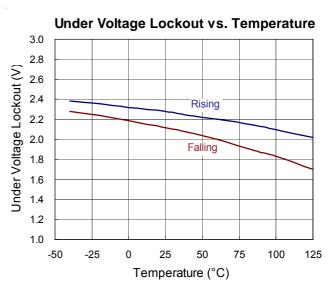






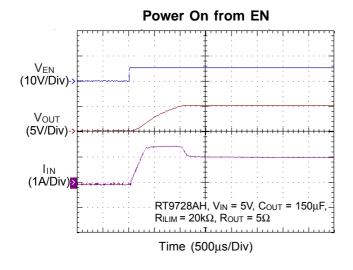


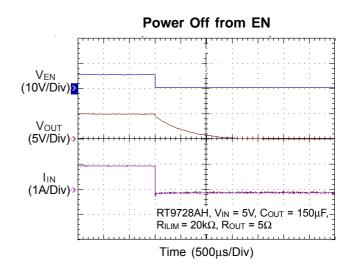


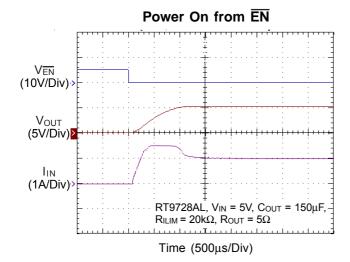


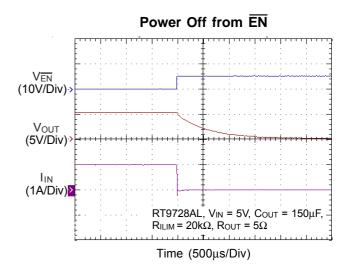
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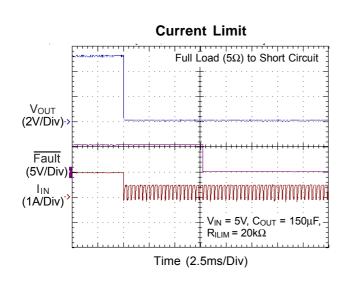


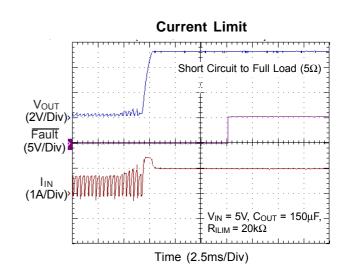








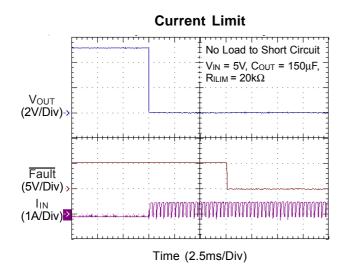


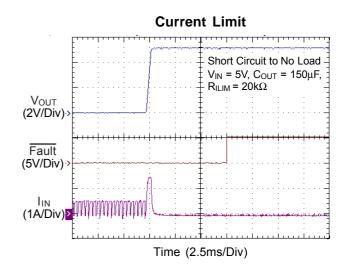


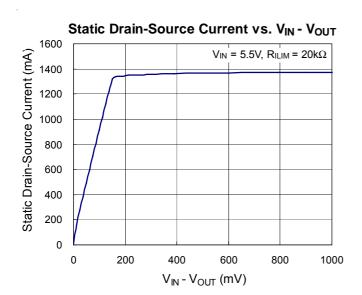
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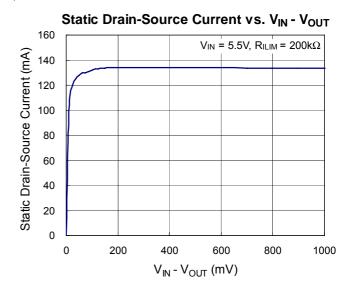
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## **Applications Information**

The RT9728A is a single P-MOSFET high side power switch with active-high/low enable input, optimized for self-powered and bus-powered Universal Serial Bus (USB) applications. The switch's low R<sub>DS(ON)</sub> meets USB voltage drop requirements and a flag output is available to indicate fault conditions to the local USB controller.

### **Current Limiting and Short Circuit Protection**

When a heavy load or short circuit situation occurs while the switch is enabled, large transient current may flow through the device. The RT9728A includes a current-limit circuitry to prevent these large currents from damaging the MOSFET switch and the hub downstream ports. The RT9728A provides an adjustable current limit threshold between 120mA and 1.3A (typ) via an external resistor,  $R_{\rm ILIM}$ , between 19.1k $\Omega$  and 232k $\Omega$ . However, if the ILIM pin is connected to  $V_{\rm IN}$ , the current limit threshold will be 75mA (typ). Once the current limit threshold is exceeded,

the device enters constant-current mode until either thermal shutdown occurs or the fault is removed. The table1 shows a recommended current limit value vs.  $R_{\rm ILIM}$ 

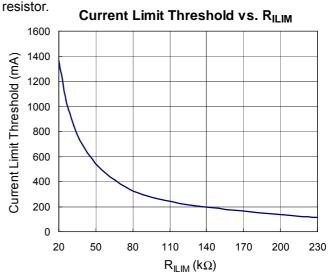


Figure 1. Current Limit Threshold vs RILIM

Table 1. Recommended RILIM Resistor Selections

Desired Nominal	Ideal Resistor	Closet 1%	Actual Limits (Include R Tolerance)			
Current Limit (mA)	$(k\Omega)$	Resistor (k $\Omega$ )	IOS Min (mA)	IOS Nom (mA)	IOS Max (mA)	
75	Short ILIM to VIN		50.0	75.0	100.0	
120	226.1	226.0	101.3	120.0	142.1	
200	134.0	133.0	173.7	201.5	233.9	
300	88.5	88.7	262.1	299.4	342.3	
400	65.9	66.5	351.1	396.7	448.7	
500	52.5	52.3	443.9	501.6	562.4	
600	43.5	43.2	535.1	604.6	674.1	
700	37.2	37.4	616.0	696.0	776.0	
800	32.4	32.4	708.7	800.8	892.9	
900	28.7	28.7	797.8	901.5	1005.2	
1000	25.8	26.1	875.4	989.1	1102.8	
1100	23.4	23.2	982.1	1109.7	1237.3	
1200	21.4	21.5	1057.9	1195.4	1332.9	
1300	19.7	19.6	1158.0	1308.5	1459.0	

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### **Fault Flag**

The RT9728A provides a FAULT signal pin which is an Nchannel open drain MOSFET output. This open drain output goes low when current exceeds current limit threshold,  $V_{OUT} - V_{IN}$  exceeds reverse voltage trip level, or the die temperature exceeds 160°C approximately. The FAULT output is capable of sinking a 1mA load to typically 180mV above ground. The FAULT pin requires a pull-up resistor; this resistor should be large in value to reduce energy drain. A  $100k\Omega$  pull-up resistor works well for most applications. In case of an over current condition, FAULT will be asserted only after the flag response delay time, tD, has elapsed. This ensures that FAULT is asserted upon valid over current conditions and that erroneous error reporting is eliminated. For example, false over current conditions may occur during hot-plug events when extremely large capacitive loads are connected, which induces a high transient inrush current that exceeds the current limit threshold. The FAULT response delay time, t<sub>D</sub>, is typically 7.5ms.

### **Supply Filter/Bypass Capacitor**

A 10µF low ESR ceramic capacitor connected from V<sub>IN</sub> to GND and located close to the device is strongly recommended to prevent input voltage drooping during hotplug events. However, higher capacitor values may be used to further reduce the voltage droop on the input. Without this bypass capacitor, an output short may cause sufficient ringing on the input (from source lead inductance) to destroy the internal control circuitry. Note that the input transient voltage must never exceed 6V as stated in the Absolute Maximum Ratings.

#### **Output Filter Capacitor**

A low ESR 150µF aluminum electrolytic capacitor connected between V<sub>OUT</sub> and GND is strongly recommended to meet the USB standard maximum droop requirement for the hub, VBUS. Standard bypass methods should be used to minimize inductance and resistance between the bypass capacitor and the downstream connector to reduce EMI and decouple voltage droop caused by hot-insertion transients in downstream cables. Ferrite beads in series with VBUS, the ground line and the 0.1µF bypass capacitors at the power connector pins are recommended for EMI and ESD protection. The bypass capacitor itself should have a low dissipation factor to allow decoupling at higher frequencies.

### **Chip Enable Input**

The RT9728AH/L will be disabled when the  $EN/\overline{EN}$  pin is in a logic-low/high condition. During this condition, the internal circuitry and MOSFET are turned off, reducing the supply current to 1µA typical. The maximum guaranteed voltage for a logic-low at the EN/EN pin is 0.66V. A minimum guaranteed voltage of 1.1V at the EN/ EN pin will turn off the RT9728A. Floating the input may cause unpredictable operation. EN/EN should not be allowed to go negative with respect to GND.

### **Under Voltage Lockout**

Under voltage lockout (UVLO) prevents the MOSFET switch from turning on until input voltage exceeds approximately 2.3V. If input voltage drops below approximately 2.1V, UVLO turns off the MOSFET switch and FAULT will be asserted accordingly. The under voltage lockout detection functions only when the switch is enabled.

#### Thermal Considerations

For continuous operation, do not exceed absolute maximum junction temperature. The maximum power dissipation depends on the thermal resistance of the IC package, PCB layout, rate of surrounding airflow, and difference between junction and ambient temperature. The maximum power dissipation can be calculated by the following formula:

 $P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA}$ 

where  $T_{J(MAX)}$  is the maximum junction temperature,  $T_A$  is the ambient temperature, and  $\theta_{JA}$  is the junction to ambient thermal resistance.

For recommended operating condition specifications, the maximum junction temperature is 125°C. The junction to ambient thermal resistance,  $\theta_{JA}$ , is layout dependent. For SOT-23-6 packages, the thermal resistance,  $\theta_{JA}$ , is 250°C/ W on a standard JEDEC 51-3 single-layer thermal test board. For WDFN-6L 2x2 packages, the thermal resistance,  $\theta_{JA}$ , is 165°C/W on a standard JEDEC 51-3

single-layer thermal test board. The maximum power dissipation at  $T_A$  = 25°C can be calculated by the following formula :

 $P_{D(MAX)}$  = (125°C - 25°C) / (250°C/W) = 0.400W for SOT-23-6 package

 $P_{D(MAX)}$  = (125°C - 25°C) / (165°C/W) = 0.606W for WDFN-6L 2x2 package

The maximum power dissipation depends on the operating ambient temperature for fixed  $T_{J(MAX)}$  and thermal resistance,  $\theta_{JA}$ . The derating curves in Figure 2 allow the designer to see the effect of rising ambient temperature on the maximum power dissipation.

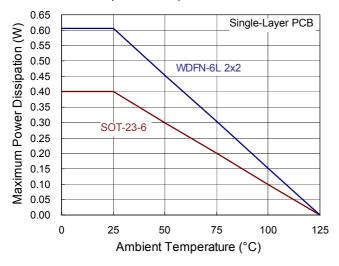
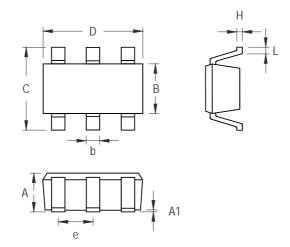


Figure 2. Derating Curve of Maximum Power Dissipation



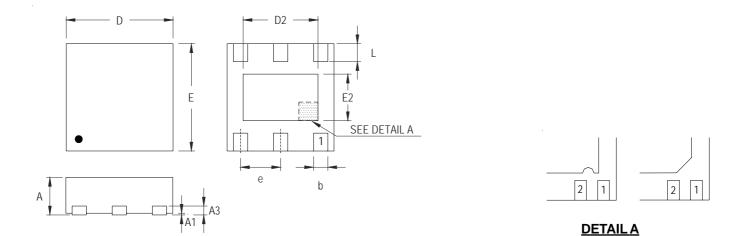
# **Outline Dimension**



Cumbal	Dimensions	n Millimeters	Dimensions In Inches		
Symbol	Min	Max	Min	Max	
А	0.889	1.295	0.031	0.051	
A1	0.000	0.152	0.000	0.006	
В	1.397	1.803	0.055	0.071	
b	0.250	0.560	0.010	0.022	
С	2.591	2.997	0.102	0.118	
D	2.692	3.099	0.106	0.122	
е	0.838	1.041	0.033	0.041	
Н	0.080	0.254	0.003	0.010	
L	0.300	0.610	0.012	0.024	

**SOT-23-6 Surface Mount Package** 





Note: The configuration of the Pin#1 identifier is optional, but must be located within the zone indicated.

Pin #1 ID and Tie Bar Mark Options

Cumbal	Dimensions	In Millimeters	Dimensions In Inches		
Symbol	Min	Max	Min	Max	
А	0.700	0.800	0.028	0.031	
A1	0.000	0.050	0.000	0.002	
A3	0.175	0.250	0.007	0.010	
b	0.200	0.350	0.008	0.014	
D	1.950	2.050	0.077	0.081	
D2	1.000	1.450	0.039	0.057	
Е	1.950	2.050	0.077	0.081	
E2	0.500	0.850	0.020	0.033	
е	0.650		0.0	)26	
L	0.300	0.400	0.012	0.016	

W-Type 6L DFN 2x2 Package

### **Richtek Technology Corporation**

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