

6MBP35XSD060-50

IGBT Modules

IGBT MODULE (X series) 600V / 35A / IPM

Features

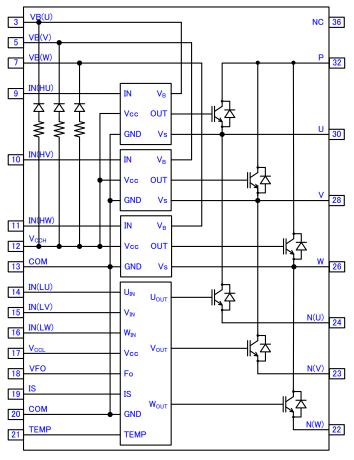
Low-side IGBTs are separate emitter type Short circuit protection Temperature sensor output function Under voltage protection Fault signal output function Input interface: TTL (3.3V/5V) Active high logic

Applications

AC 100 ~ 240V three phase inverter drive for small power AC motor drives (such as compressor motor drive for air conditioner, compressor motor drive for heat pump applications, fan motor drive, ventilator motor drive)



Terminal assign and Internal circuit



Pin No.	Pin Name	Pin Description
3	VB(U)	High-side bias voltage for U-phase IGBT driving
5	VB(V)	High-side bias voltage for V-phase IGBT driving
7	VB(W)	High-side bias voltage for W-phase IGBT driving
9	IN(HU)	Signal input for high side U-phase
10	IN(HV)	Signal input for high side V-phase
11	IN(HW)	Signal input for high side W-phase
12	Vссн	High-side control supply
13	СОМ	Common supply ground
14	IN(LU)	Signal input for low side U-phase
15	IN(LV)	Signal input for low side V-phase
16	IN(LW)	Signal input for low side W-phase
17	Vccl	Low-side control supply
18	VFO	Fault output
19	IS	Over current sensing voltage input
20	COM	Common supply ground
21	TEMP	Temperature sensor output
22	N(W)	Negative bus voltage input for W-phase
23	N(V)	Negative bus voltage input for V-phase
24	N(U)	Negative bus voltage input for U-phase
26	W	Motor W-phase output
28	V	Motor V-phase output
30	U	Motor U-phase output
32	Р	Positive bus voltage input
36	NC	No Connection

■ Absolute Maximum Ratings at T_j=25°C, V_{cc}=15V (unless otherwise specified)

Ite	ems	Symbol	Characteristics	Unit	Remarks
	DC Bus Voltage	V _{DC}	450	V	Note *1
	Bus Voltage (Surge)	V _{DC(Surge)}	500	V	Note *1
	Collector-Emitter Voltage	Vces	600	V	
	Collector Current	Ic@25	35	Α	Note *2
ock	Deals Calleston Comment		70	А	Vcc≧15V, Vв(*)≧15V Note *2, *3, *4
nverter block	Peak Collector Current	CP@25	60	А	Vcc≧13V, Vв(*)≧13V Note *2, *3, *4
nVe	Diode Forward current	I _{F@25}	35	Α	Note *2
_	Peak Diode Forward current	IFP@25	70	Α	Note *2
	Collector Power Dissipation	P _{D_IGBT}	79.1	W	per single IGBT Tc=25°C
	FWD Power Dissipation	P _{D_FWD}	39.6	W	per single FWD Tc=25°C
	Junction Temperature	T _i	150	°C	
	Operating Junction Temperature	T _{jOP}	-40 ~ +150	°C	
	High-side Supply Voltage	Vссн	-0.5 ~ 20	V	Applied between Vcch-COM
	Low-side Supply Voltage	Vccl	-0.5 ~ 20	V	Applied between Vccl-COM
	High-side Bias Absolute Voltage	Vvb(u)-com Vvb(v)-com Vvb(w)-com	-0.5 ~ 620	V	Applied between VB(U)-COM, VB(W)-COM
t block		$\begin{matrix} V_{B(U)} \\ V_{B(V)} \\ V_{B(W)} \end{matrix}$	-0.5 ~ 20	V	Note *4
ol circuit	_	Vu Vv Vw	-5 ~ 600	V	Applied between U-COM, V-COM, W-COM Note *5
Control	Input Signal Voltage	Vin	-0.5 ~ V _{CCH} +0.5 -0.5 ~ V _{CCL} +0.5	V	Note *6
	Input Signal Current	lin	3	mA	sink current
	Fault Signal Voltage	V _{FO}	-0.5 ~ V _{ссн} +0.5	V	Applied between V _F O-COM
	Fault Signal Current	IFO	1	mA	sink current
	Over Current sensing Input Voltage	Vis	-0.5 ~ V _{CCL} +0.5	V	Applied between IS-COM
	Junction Temperature	T _j	150	°C	
	perating Case Temperature	Tc	-40 ~ +125	°C	See Fig.1-1
St	orage Temperature	T _{stg}	-40 ~ +125	°C	
Is	olation Voltage	Viso	AC 1500	Vrms	Sine wave,60Hz t=1min , Note *7

Note *1 : Applied between P-N(U),P-N(V),P-N(W)
Note *2 : Pulse width and duty were limited by T_imax.
Note *3 : V_C is applied between V_{COH}-COM,V_{CCL}-COM.
Note *4 : V_B(*) is applied between VB(U)-U,VB(V)-V, VB(W)-W.
Note *5 : Over 13.0V applied between VB(U)-U,VB(V)-V, VB(W)-W. This IPM module might make incorrect response if the high-side bias offset voltage is less than -5V.

Note *6 : Applied between IN(HU)-COM,IN(HV)-COM,IN(HW)-COM,IN(LU)-COM,IN(LV)-COM,IN(LW)-COM. Note *7 : Applied between shorted all terminal and IMS (Insulated Metal Substrate).

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■ Electrical Characteristics

● Inverter block (Tj=25°C unless otherwise specified)

Description	Symbol	Conditions		min.	typ.	max.	Unit
Zero gate Voltage	Ices	V _{CE} = 600V	T _i =25°C	-	-	1	mA
Collector current	ICES	V _{IN} = 0V	T _j =125°C	-	-	10	mA
		V _{cc} =+15V	Ic=3.5A T _j =25°C	-	0.80	1.00	
Collector-Emitter saturation Voltage	V _{CE(sat)}	V _B (*)=+15V V _{IN} =5V	Ic=35A T _j =25°C	-	1.40	1.70	V
		Note *4	Ic=35A T _j =125°C	-	1.55	1.90	
EMD Forward voltons dues	VF	I _F =35A	T _j =25°C	-	1.70	2.05	V
FWD Forward voltage drop	VF	V _{IN} =0V	T _j =125°C	-	1.55	-	
Turn-on time	ton		·	0.54	0.95	1.35	
Turn-on delay	t _{d(on)}	Vpc=300V			0.80	-	
Turn-on rise time	t _r	Ic=35A		-	0.15	-	
VCE-IC Cross time of turn-on	t _{c(on)}	V _{cc} =15V		-	0.30	0.50	
Turn-off time	toff	│ V _B (*)=15V │ T _i =125°C		-	1.35	1.85	μs
Turn-off delay	t _{d(off)}	t_{doff} $V_{\text{IN}}=0$ V <-> 5V See Fig.2-1 Note *4		-	1.15	-	
Turn-off fall time	t _f			-	0.10	-	
VCE-IC Cross time of turn-on	t _{c(off)}			-	0.10	0.20	
FWD Reverse Recovery time	trr			-	0.20	-	

● Control circuit block (T_j=25°C unless otherwise specified)

Description	Symbol	Conditions		min.	typ.	max.	Unit
Circuit current of Low-side		V _{CCL} =15V	V _{IN} =5V	-	0.6	0.9	mA
Circuit current of Low-side	ICCL	V _{CCL} =15V	V _{IN} =0V	-	0.6	0.9	IIIA
Civavit accurant of High aids		V _{CCH} =15V	V _{IN} =5V	-	1.25	1.9	mA
Circuit current of High-side	Іссн	V _{CCH} =15V	V _{IN} =0V	-	1.25	1.9	MA
Circuit aureant of Pastatran aircuit (nor and uint)		V _B (U)=15V,	V _{IN} =5V	-	-	0.20	mA
Circuit current of Bootstrap circuit (per one uint	Ісснв	V _B (V)=15V, V _B (W)=15V	V _{IN} =0V	-	-	0.20	IIIA
Input Signal threshold valtage	Vth _(on)	Note *8 Pw≥0.9μs		-	2.1	2.6	V
Input Signal threshold voltage	Vth _(off)			0.8	1.3	-	V
Input Signal threshold hysteresis voltage	Vth _(hys)			0.35	0.80	-	V
Operational input pulse width of turn-on	t _{IN(ON)}	V _{IN} =0V to 5V rise up, Note *6, Note *8		0.5	-	-	μs
Operational input pulse width of turn-off	t _{IN(OFF)}	V _{IN} =5V to 0V fall down Note *6, Note *8		1.0	-	-	μs
Input current	lin	V _{IN} =5V, Note *6		0.7	1.0	1.5	mA
Input pull-down resistance	Rin	Note *6		3.3	5.0	7.2	kΩ
Fault Output Voltage	V _{FO(H)}	V_{is} =0V, VFO terminal pull up to 5V by $10k\Omega$		4.9	-	-	V
	V _{FO(L)}	V _{IS} =1V, I _{FO} =1mA		-	-	0.95	V
Fault Output pulse width	t FO	Note *9, See Fig.2-2, 2-3		20	-	-	μs

Control circuit block (continued)

Over Current Protection Voltage Level	V _{IS(ref)} V _{CC} =15V		0.455	0.48	0.505	V	
Over Current Protection Delay time	td _(IS)	─ Note *3, *10 See Fig.2-2		0.3	0.8	1.3	μs
Output Voltage of temperature sensor	V _(temp)	Note *44	Tc=90°C	2.63	2.77	2.91	V
		Note *11	Tc=25°C	0.88	1.13	1.39	V
Vcc Under Voltage Trip Level of Low-side	V _{CCL(OFF)}			10.3	-	12.5	V
Vcc Under Voltage Reset Level of Low-side	V _{CCL(ON)}	See Fig.2-3		10.8	-	13.0	V
Vcc Under Voltage hysteresis	V _{CCL(hys)}			-	0.5	-	V
Vcc Under Voltage Trip Level of High-side	V _{CCH(OFF)}	T _I <150°C See Fig.2-4		8.3	-	10.3	V
Vcc Under Voltage Reset Level of High-side	V _{CCH(ON)}			8.8	-	10.8	V
Vcc Under Voltage hysteresis	V _{CCH(hys)}			-	0.5	-	V
VB Under Voltage Trip Level	V _{B(OFF)}	T _j <150°C See Fig.2-5		10.0	-	12.0	V
VB Under Voltage Reset Level	V _{B(ON)}			10.5	-	12.5	V
VB Under Voltage hysteresis	V _{B(hys)}			-	0.5	-	V
Famus and scale from the first of the state	V _{F(BSD)}	T _j =25°C I _{F(BSD)} =10mA		0.90	1.4	1.90	V
Forward voltage of Bootstrap diode	V _{F(BSD)}	T _j =25°C I _{F(BSD)} =100mA		2.3	4.3	6.3	V

Note *8: This IPM module might make incorrect response if the input signal pulse width is less than t_{IN(ort)} and t_{IN(ort)}.

Note *9: Fault signal is asserted corresponding to an "Over-current protection", an "Under-voltage protection" at low-side, and an "Over-heat protection".

Under the condition of "Over-current protection" or "Under-voltage protection" or "Over-heat protection", the fault signal is asserted continuously while these conditions are continuing. However, the minimum fault output pulse width is minimum 20µsec even if very short failure condition (which is less than 20µs) is triggered.

Note *10: Over current protection is functioning only for the low-side arms. Note *11: Fig.1-1 shows the measurement position of temperature sensor.

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■ Thermal Characteristics

Description	Symbol	min.	typ.	max.	Unit
Junction to Case Thermal Resistance (per single IGBT) Note *12	Rth(j-c)_IGBT	-	-	1.58	°C/W
Junction to Case Thermal Resistance (per single FWD) Note *12	Rth(j-c)_FWD	-	-	3.16	°C/W

Note *12: Thermal compound with good thermal conductivity should be applied evenly with about +100µm~+200µm on the contacting surface of this device and heat-

■ Mechanical Characteristics

Description	Symbol	Conditions	min.	typ.	max.	Unit
Tighten torque	-	Mounting screw: M3	0.59	0.69	0.98	Nm
Heat-sink side flatness	-	Note. *13	0	-	100	μm
Weight	-	-	-	9.3	-	g

Note *13: Fig.1-2 shows the measurement position of heat sink flatness

■ Recommended Operation Conditions

● All voltages are absolute voltages referenced to Vcc -potential unless otherwise specified.

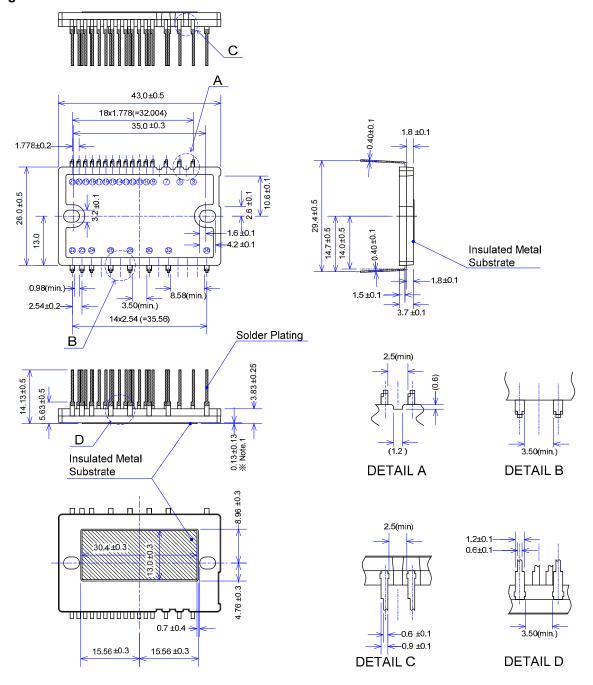
Description	Conditions	min.	typ.	max.	Unit
DC Bus Voltage	V _{DC}	0	300	400	V
High-side Bias Voltage for IGBT gate driving	V _B (*)	13.0	15.0	18.5	V
High-side Supply Voltage	Vccн	13.5	15.0	16.5	V
Low-side Supply Voltage	VccL	13.5	15.0	16.5	V
Control Supply variation	ΔV _B	-1	-	1	\//uo
Control Supply variation	ΔVcc	-1	-	1	V/µs
Input signal voltage	V _{IN}	0	-	5	V
Voltage for current sensing	Visc	0	-	5	V
Potential difference of between COM to N (including surge)	V _{COM_N}	-5	-	5	V
Dead time for preventing arm-short (Tc≤125°C)	toead	1.0	-	-	μs
Allowable output current (Note *14)	lo	-	-	35.0	A rms
Allowable minimum input pulse width	PW _{IN(on)}	0.5	-	-	μs
(Note *15, Note *16)	PW _{IN(off)}	1.0	-	-	μs
PWM Input frequency	f _{PWM}	-	-	20	kHz
Operating Junction Temperature	T _{j(ope)}	-30	-	150	°C

Note *14: V_{DC}=300V, V_{CCH}=V_{CCL}=V_B(*)=15V, PF=0.8, Sinusoidal PWM, 3phase modulation, T_J≤150°C, T_C≤100°C, f_{PWM}=5kHz, fo=200Hz, Ks=0.9 Note *15: In the pulse width of 0.5us, the loss of IGBT increases for the saturation operation.

To reduce the loss of IGBT, please enlarge the pulse width more than the switching time of IGBT.

Note *16: This IPM module might response according to input signal pulse even when the input signal pulse width is less than PW_{IN(orl)} and PW_{IN(orl)}.

■ Package outline dimensions



Unit: mm

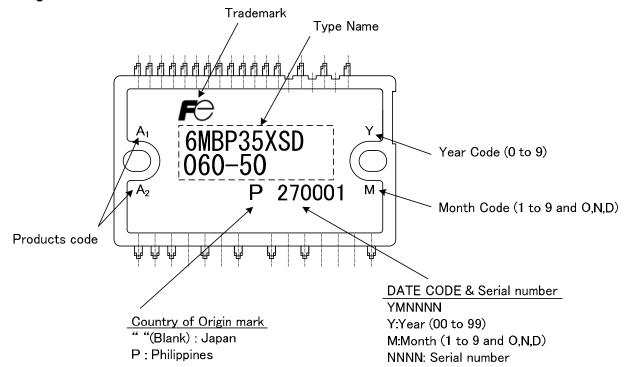
Note %1: The IMS(Insulated Metal Substrate) deliberately protruded from back surface of case. It is improved of thermal conductivity between IMS and heat-sink.

Pin No.	Pin Name
3	VB(U)
5	VB(V)
7	VB(W)
9	IN(HU)
10	IN(HV)
11	IN(HW)
12	Vссн
13	COM

Pin No.	Pin Name
14	IN(LU)
15	IN(LV)
16	IN(LW)
17	V _{CCL}
18	VFO
19	IS
20	СОМ
21	TEMP

Pin No.	Pin Name
22	N(W)
23	N(V)
24	N(U)
26	W
28	V
30	U
32	Р
36	NC

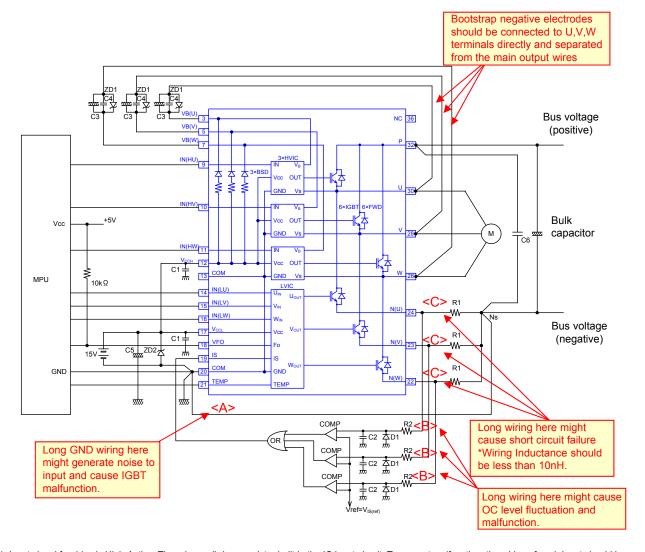
Marking



Note : Product code A_1 means current ratings , and "P" is marked. Product code A_2 means variations , and "D" is marked.

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- An example of application circuit.
- Fig. shows an example of an application circuit.



- Note *1: Input signal for drive is High-Active. There is a pull-down resistor built in the IC input circuit. To prevent malfunction, the wiring of each input should be as short as possible. When using R-C coupling circuit, make sure the input signal level meet the turn-on and turn-off threshold voltage.
- Note *2: By the function of the HVIC, it is possible of the direct coupling to microprocessor (MPU) without any photo-coupler or pulse-transformer isolation.
- Note *3: VFO output is open drain type. It should be pulled up to the positive side of a 5V power supply by a resistor of about $10k\Omega$.
- Note *4: To prevent erroneous protection, the wiring of (A), (B), (C) should be as short as possible.
- Note *5: The time constant R2-C2 of the protection circuit should be selected approximately 1.5µs.
 - Over current (OC) shutdown time might vary due to the wiring pattern. Tight tolerance, temp-compensated type is recommended for R2, C2.
- Note *6: Please set the threshold voltage of the comparator reference input to be same as the IPM OC trip reference voltage V_{IS(ref)}.
- Note *7: Please use high speed type comparator and logic IC to detect OC condition quickly.
- Note *8: If negative voltage of R1 at the switching timing is applied, the schottky barrier diode D1 is recommended to be inserted parallel to R1.
- Note *9: All capacitors should be mounted as close to the terminals of the IPM as possible. (C1, C4: narrow temperature drift, higher frequency and DC bias characteristic ceramic type are recommended, and C3, C5: narrow temperature drift, higher frequency and electrolytic type.)
- Note *10: To prevent surge destruction, the wiring between the snubber capacitor and the P terminal ,Ns node should be as short as possible. Generally a 0.1μ to 0.22μF snubber capacitor (C6) between the P terminal and Ns node is recommended.
- Note *11: Two COM terminals (13 & 20 pin) are connected inside the IPM, it must be connected either one to the signal GND outside and leave another one open.
- Note *12: It is recommended to insert a zener-diode (22V) between each pair of control supply terminals to prevent surge destruction.
- Note *13: If signal GND is connected to power GND by broad pattern, it may cause malfunction by power GND fluctuation. It is recommended to connect signal GND and power GND at only a point.

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Fig.1-2: The measurement position of heat sink flatness

Fig.1-1: The measurement position of temperature sensor.

Temperature sensor position

Approx. 7.0

Approx. 6.3

Heat sink side

Tc measurement position

SIDE VIEW

TOP VIEW

Fig.2-1: Switching waveforms

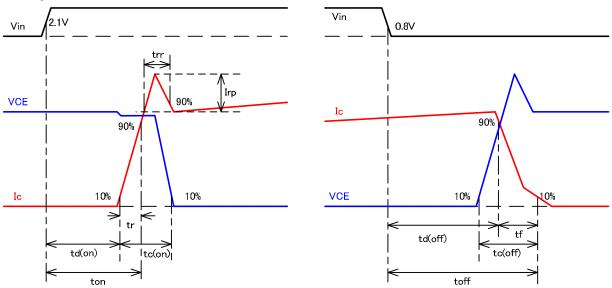
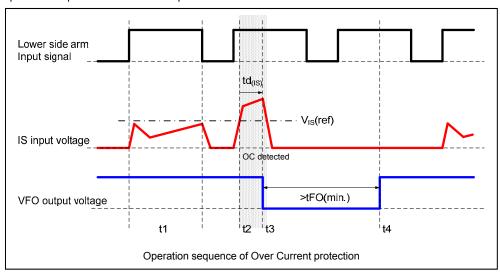
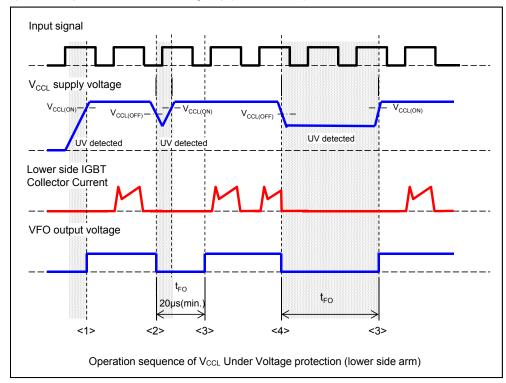


Fig.2-2: Operation sequence of Over current protection



- t1: IS input voltage does not exceed Vis(ref), while the collector current of the lower side IGBT is under the normal operation.
- t2: When IS input voltage exceeds Vis(ref), the OC is detected.
- t3: The fault output VFO is activated and all lower side IGBT shut down simultaneously after the over current protection delay time td(IS). Inherently there is dead time of LVIC in td(IS)
- t4: After the fault output pulse width tFO, the OC is reset. Then next input signal is activated.

Fig.2-3: Operation sequence of $V_{\text{\tiny CCL}}$ Under voltage trip (lower side arm)



When VccL is under 4V, UV and fault output are not activated.

- When VccL is under VccLoN, all lower side IGBTs are OFF state.

 After VccL rises VccLoN, the fault output VFO is released (high level).

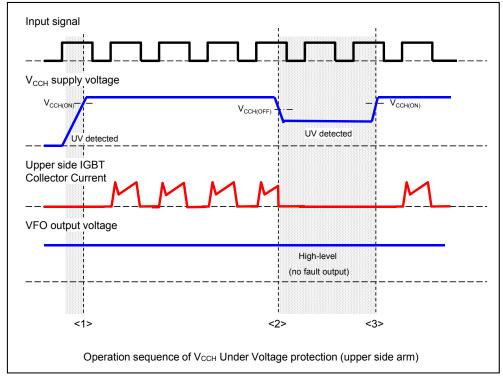
 And the LVIC starts to operate, then next input is activated.

 42> The fault output VFO is activated when VccL falls below VccLoFF), and all lower side IGBT remains OFF state. When the voltage drop time is less than 20µs, the fault output pulse width is generated minimum 20µs and all lower side IGBTs are OFF state in spite of input signal condition during that time.
- <3> UV is reset after the when VccL exceeds VccL(ON) and the fault output VFO is reset simultaneously.
- And the LVIC starts to operate, then next input is activated.

 <4> When the voltage drop time is more than t_{FO}, the fault output pulse width is generated and all lower side IGBTs are OFF state in spite of input signal condition during the same time.

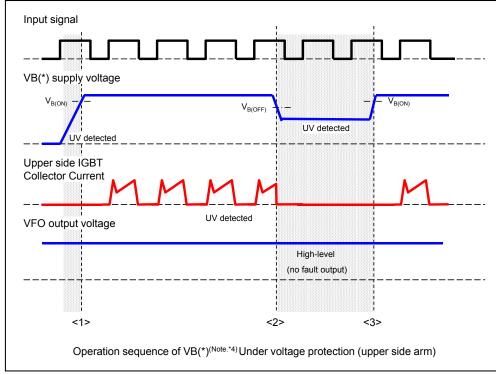
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Fig.2-4: Operation sequence of V_{CCH} Under voltage trip (upper side arm)



- <1> When Vccн is under Vccн(ом), the upper side IGBT is OFF state.
 After Vccн exceeds Vccн(ом), the HVIC starts to operate. Then next input is activated.
 The fault output VFO is constant (high level) not to depend on Vccн.
- <2> After Vcch falls below VcchoFF, the upper side IGBT remains OFF state. But the fault output VFO keeps high level.
- <3> The HVIC starts to operate after UV is reset, then next input is activated.

Fig.2-5: Operation sequence of VB Under voltage trip (upper side arm)



<1> When VB(*) is under $V_{B(ON)}$, the upper side IGBT is OFF state. After VB(*) exceeds $V_{B(ON)}$, the HVIC starts to operate. Then next input is activated. The fault output VFO is constant (high level) not to depend on VB(*). (Note*14)

Note *14: The fault output is not given HVIC bias conditions.

<2> After VB(*) falls below V_{B(OFF)}, the upper side IGBT remains OFF state. But the fault output VFO keeps high level.

<3> The HVIC starts to operate after UV is reset, then next input is activated.

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