

6MBP20XSF060-50

IGBT Modules

IGBT MODULE (X series) 600V / 20A / IPM

Features

Low-side IGBTs are separate emitter type Short circuit protection Temperature sensor output function Overheating protection Under voltage protection Fault signal output function

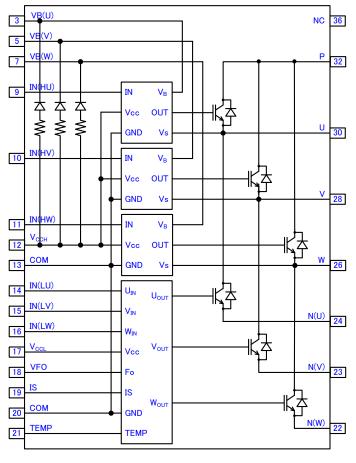
Input interface: TTL (3.3V/5V) Active high logic

Applications

AC 100 ~ 240V three phase inverter drive for small power AC motor drives (such as compressor motor drive for air conditioner, compressor motor drive for heat pump applications, fan motor drive, ventilator motor drive)



■ Terminal assign and Internal circuit



| Pin No. | Pin Name | Pin Description |
|---------|------------------|---|
| 3 | VB (U) | High-side bias voltage for U-phase IGBT driving |
| 5 | VB (V) | High-side bias voltage for V-phase IGBT driving |
| 7 | VB (W) | High-side bias voltage for W-phase IGBT driving |
| 9 | IN (HU) | Signal input for high side U-phase |
| 10 | IN (HV) | Signal input for high side V-phase |
| 11 | IN (HW) | Signal input for high side W-phase |
| 12 | V _{ссн} | High-side control supply |
| 13 | COM | Common supply ground |
| 14 | IN (LU) | Signal input for low side U-phase |
| 15 | IN (LV) | Signal input for low side V-phase |
| 16 | IN(LW) | Signal input for low side W-phase |
| 17 | Vccl | Low-side control supply |
| 18 | VFO | Fault output |
| 19 | IS | Over current sensing voltage input |
| 20 | COM | Common supply ground |
| 21 | TEMP | Temperature sensor output |
| 22 | N (W) | Negative bus voltage input for W-phase |
| 23 | N (V) | Negative bus voltage input for V-phase |
| 24 | N (U) | Negative bus voltage input for U-phase |
| 26 | W | Motor W-phase output |
| 28 | V | Motor V-phase output |
| 30 | U | Motor U-phase output |
| 32 | Р | Positive bus voltage input |
| 36 | NC | No Connection |

■ Absolute Maximum Ratings at T_j=25°C, V_{cc}=15V (unless otherwise specified)

| Ite | ems | Symbol | Characteristics | Unit | Remarks |
|---------------|--|---|--|------|--|
| | DC Bus Voltage | V _{DC} | 450 | V | Note *1 |
| | Bus Voltage (Surge) | V _{DC(Surge)} | 500 | V | Note *1 |
| | Collector-Emitter Voltage | Vces | 600 | V | |
| | Collector Current | Ic@25 | 20 | Α | Note *2 |
| ock | Peak Collector Current | | 40 | A | Vcc≧15V, V _B (*)≧15V Note *2, *3, *4 |
| nverter block | reak collector current | ICP@25 | 30 | A | Vcc≧13V, Vв(*)≧13V Note *2, *3, *4 |
| Ž | Diode Forward current | I _{F@25} | 20 | Α | Note *2 |
| _ | Peak Diode Forward current | FP@25 | 40 | Α | Note *2 |
| | Collector Power Dissipation | P _{D_IGBT} | 41.0 | W | per single IGBT T₀=25°C |
| | FWD Power Dissipation | P _{D_FWD} | 27.8 | W | per single FWD T _c =25°C |
| | Junction Temperature | Tj | 150 | °C | |
| | Operating Junction Temperature | Тјор | -40 ~ +150 | °C | |
| | High-side Supply Voltage | V _{ссн} | -0.5 ~ 20 | V | Applied between Vcch-COM |
| | Low-side Supply Voltage | Vccl | -0.5 ~ 20 | V | Applied between Vccl-COM |
| | High-side Bias Absolute Voltage | Vvb(u)-com Vvb(v)-com Vvb(w)-com | -0.5 ~ 620 | V | Applied between VB(U)-COM, VB(V)-COM, VB(W)-COM |
| t block | High-side Bias Voltage for IGBT gate driving | V _{B(U)} V _{B(V)} V _{B(W)} | -0.5 ~ 20 | V | Note *4 |
| ol circuit | | Vu Vv Vw | -5 ~ 600 | V | Applied between U-COM, V-COM, W-COM Note *5 |
| Control | Input Signal Voltage | Vin | -0.5 ~ V _{ссн} +0.5 -0.5 ~ V _{ссг} +0.5 | V | Note *6 |
| | Input Signal Current | I _{IN} | 3 | mA | sink current |
| | Fault Signal Voltage | V _{FO} | -0.5 ~ Vccl+0.5 | V | Applied between VFO-COM |
| | Fault Signal Current | I FO | 1 | mA | sink current |
| | Over Current sensing Input Voltage | Vis | -0.5 ~ V _{CCL} +0.5 | V | Applied between IS-COM |
| | Operating Junction Temperature | Tj | 150 | °C | |
| _ | perating Case Temperature | Tc | -40 ~ +125 | °C | See Fig.1-1 |
| St | orage Temperature | T _{stg} | -40 ~ +125 | °C | |
| Is | olation Voltage | Viso | AC 1500 | Vrms | Sine wave, 60Hz t=1min, Note *7 |

Note *1 : Applied between P-N(U), P-N(V), P-N(W)
Note *2 : Pulse width and duty were limited by Tjmax.
Note *3 : Vcc is applied between Vcc+-COM, Vcc+-COM.
Note *4 : Ve(*) is applied between VB(U)-U, VB(V)-V, VB(W)-W.
Note *5 : Over 13.0V applied between VB(U)-U, VB(V)-V, VB(W)-W. This IPM module might make incorrect response if the high-side bias offset voltage is less than -5V.

Note *6 : Applied between IN(HU)-COM, IN(HV)-COM, IN(HW)-COM, IN(LU)-COM, IN(LV)-COM, IN(LV)-COM. Note *7 : Applied between shorted all terminal and IMS (Insulated Metal Substrate).

■ Electrical Characteristics

● Inverter block (Tj=25°C unless otherwise specified)

| Description | Symbol | Conditions | | min. | typ. | max. | Unit |
|--------------------------------------|----------------------|--|--|------|------|------|------|
| Zero gate Voltage | Ices | Vce=600V | T _j =25°C | - | - | 1 | mA |
| Collector current | ICES | V _{IN} =0V | T _i =125°C | - | - | 10 | mA |
| | | Vcc=+15V | Ic=2A T _j =25°C | - | 0.90 | 1.10 | |
| Collector-Emitter saturation Voltage | V _{CE(sat)} | V _B (*)=+15V V _{IN} =5V | I _c =20A T _j =125°C | - | 1.60 | 1.90 | V |
| | | Note *4 | Ic=20A T _j =125°C | - | 1.75 | 2.10 | |
| FMD Formulation during | V _F | I _F =20A V _{IN} =0V | T _j =25°C | - | 1.70 | 2.05 | V |
| FWD Forward voltage drop | | | T _j =125°C | 0.60 | 1.50 | - | |
| Turn-on time | ton | | | - | 1.00 | 1.35 | |
| Turn-on delay | t _{d(on)} | V _{DC} =300V | | - | 0.80 | - | |
| Turn-on rise time | t _r | Ic=20A | | - | 0.20 | - | |
| VCE-IC Cross time of turn-on | t _{c(on)} | Vcc=15V | | - | 0.40 | 0.65 | |
| Turn-off time | toff | │ V _B (*)=15V │ T _i =125°C | | - | 1.00 | 1.40 | μs |
| Turn-off delay | t _{d(off)} | V _{IN} =0V <-> 5V See Fig.2-1 Note *4 | | - | 0.90 | - | |
| Turn-off fall time | t _f | | | - | 0.10 | - | |
| VCE-IC Cross time of turn-on | t _{c(off)} | | | - | 0.15 | 0.30 | |
| FWD Reverse Recovery time | trr | | | | 0.20 | - | |

● Control circuit block (T_j=25°C unless otherwise specified)

| Description | Symbol | Conditions | | min. | typ. | max. | Unit |
|---|---|---|---------------------|-------|------|-------|-------|
| | | V _{CCL} =15V | V _{IN} =5V | - | 0.6 | 0.9 | |
| Circuit current of Low-side | Icc. | V _{CCL} =15V | V _{IN} =0V | - | 0.6 | 0.9 | mA |
| | | V _{CCH} =15V | V _{IN} =5V | - | 1.25 | 1.9 | |
| Circuit current of High-side | Іссн | V _{CCH} =15V | V _{IN} =0V | - | 1.25 | 1.9 | mA |
| | | V _в (U)=15V, | V _{IN} =5V | - | _ | 0.20 | |
| Circuit current of Bootstrap circuit (per one uint) | Ісснв | V _B (V)=15V, | V _{IN} =0V | _ | | 0.20 | mA |
| | \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ | V _B (W)=15V | V _{IN} =UV | - | | | |
| Input Signal threshold voltage | Vth _(on) | Note *8 | | - | 2.1 | 2.6 | V |
| In most Cinnal threath and houstonesis walters | Vth _(off) | – Pw≥0.9μs | | 0.8 | 1.3 | - | V |
| Input Signal threshold hysteresis voltage | Vth _(hys) | V _{IN} =0V to 5V rise up | | 0.35 | 0.80 | - | V |
| Operational input pulse width of turn-on | t _{IN(ON)} | Note *6, Note *8 | | 0.5 | - | - | μs |
| On a mation of the most market wildth of turns off | | V _{IN} =5V to 0V fall down | | 0.7 | | | |
| Operational input pulse width of turn-off | t _{IN(OFF)} | Note *6, Note *8 | | 0.7 | - | - | μs |
| Input current | In | V _{IN} =5V, Note *6 | | 0.7 | 1.0 | 1.5 | mA |
| Input pull-down resistance | Rin | Note *6 | | 3.3 | 5.0 | 7.2 | kΩ |
| Fault Output Voltage | V _{FO(H)} | V _{IS} =0V, VFO terminal pull ι 10kΩ | ıp to 5V by | 4.9 | - | - | V |
| | V _{FO(L)} | V _{IS} =1V, I _{FO} =1mA | | - | - | 0.95 | V |
| Fault Output pulse width | t FO | Note *9, See Fig.2-2, 2-3 | | 20 | - | - | μs |
| Over Current Protection Voltage Level | V _{IS(ref)} | V _{cc} =15V Note *3, *10 See Fig.2-2 | | 0.455 | 0.48 | 0.505 | V |
| Over Current Protection Delay time | td _(IS) | | | 0.3 | 8.0 | 1.3 | μs |
| Output Voltage of temperature sensor | V _(temp) | Note *11 | Tc=90°C | 2.63 | 2.77 | 2.91 | V |
| Output voitage of temperature sensor | V (temp) | Note 11 | Tc=25°C | 0.88 | 1.13 | 1.39 | V |
| LVIC Overheating protection | Тон | See Fig.2-6 | | 136 | 143 | 150 | °C |
| Тон Hysteresis | T _{OH(hys)} | 366 r ig.2-0 | | 4 | 10 | 20 | °C |
| Vcc Under Voltage Trip Level of Low-side | V _{CCL(OFF)} | T 4450°C | | 10.3 | - | 12.5 | V |
| Vcc Under Voltage Reset Level of Low-side | V _{CCL(ON)} | T _i <150°C See Fig.2-3 | | 10.8 | - | 13.0 | V |
| Vcc Under Voltage hysteresis | V _{CCL(hys)} | | | - | 0.5 | - | V |
| Vcc Under Voltage Trip Level of High-side | V _{CCH(OFF)} | T :45000 | | 8.3 | - | 10.3 | V |
| Vcc Under Voltage Reset Level of High-side | V _{CCH(ON)} | ⊟ T _i <150°C – See Fig.2-4 | | 8.8 | - | 10.8 | V |
| Vcc Under Voltage hysteresis | V _{CCH(hys)} | ys) Jee Fig.2-4 | | - | 0.5 | - | V |
| VB Under Voltage Trip Level | $V_{\text{B(OFF)}}$ | | | 10.0 | - | 12.0 | V |
| VB Under Voltage Reset Level | V _{B(ON)} T _i <150°C See Fig.2-5 | | | 10.5 | - | 12.5 | V |
| VB Under Voltage hysteresis | V _{B(hys)} | _ | | - | 0.5 | - | V |
| Forward voltage of Pootstrap diede | V _{F(BSD)} | T _j =25°C I _{F(BSD)} =10mA | | 0.90 | 1.4 | 1.90 | V |
| Forward voltage of Bootstrap diode | V _{F(BSD)} | T _j =25°C I _{F(BSD)} =100mA | | 2.3 | 4.3 | 6.3 | v |

Note *8: This IPM module might make incorrect response if the input signal pulse width is less than t_{IN(orf)} and t_{IN(orf)}.

Note *9: Fault signal is asserted corresponding to an "Over-current protection" and an "Under-voltage protection" at low-side, and an "Over-heat protection".

Under the condition of "Over-current protection" or "Under-voltage protection", or "Over-heat protection", the fault signal is asserted continuously while these conditions are continuing. However, the minimum fault output pulse width is minimum 20µsec even if very short failure condition (which is less than 20µs) is the condition of the condition

20µs) is triggered.

Note *10 : Over current protection is functioning only for the low-side arms.

Note *11 : Fig.1-1 shows the measurement position of temperature sensor.

■ Thermal Characteristics

| Description | | Symbol | min. | typ. | max. | Unit |
|--|-----------------|---------------------------|------|------|------|------|
| Junction to Case The (per single IGBT) Note *12 | rmal Resistance | R _{th(j-c)_IGBT} | - | - | 3.05 | °C/W |
| Junction to Case The (per single FWD) Note *12 | rmal Resistance | R _{th(j-o)_FWD} | - | - | 4.50 | °C/W |

Note *12: Thermal compound with good thermal conductivity should be applied evenly with about +100µm~+200µm on the contacting surface of this device and heat-

■ Mechanical Characteristics

| Description | Symbol | Conditions | min. | typ. | max. | Unit |
|-------------------------|--------|--------------------|------|------|------|------|
| Tighten torque | - | Mounting screw: M3 | 0.59 | 0.69 | 0.98 | Nm |
| Heat-sink side flatness | - | Note. *13 | 0 | - | 100 | μm |
| Weight | - | - | - | 9.3 | - | g |

Note *13: Fig.1-2 shows the measurement position of heat sink flatness

■ Recommended Operation Conditions

● All voltages are absolute voltages referenced to Vcc -potential unless otherwise specified.

| Description | Conditions | min. | typ. | max. | Unit |
|--|-----------------------|------|------|------|--------|
| DC Bus Voltage | V _{DC} | 0 | 300 | 400 | V |
| High-side Bias Voltage for IGBT gate driving | V _B (*) | 13.0 | 15.0 | 18.5 | V |
| High-side Supply Voltage | V _{ссн} | 13.5 | 15.0 | 16.5 | V |
| Low-side Supply Voltage | Vccl | 13.5 | 15.0 | 16.5 | V |
| Control Supply variation | ΔV _B | -1 | - | 1 | 1//110 |
| Control Supply variation | ΔVcc | -1 | - | 1 | V/µs |
| Input signal voltage | Vin | 0 | - | 5 | V |
| Voltage for current sensing | Visc | 0 | - | 5 | V |
| Potential difference of between Vcc to N (including surge) | V _{CC_N} | -5 | - | 5 | V |
| Dead time for preventing arm-short (Tc≤125°C) | toead | 1.0 | - | - | μs |
| Allowable output current (Note *14) | lo | - | - | 20.0 | A rms |
| Allowable minimum input pulse width | PW _{IN(on)} | 0.5 | - | - | μs |
| (Note *15, Note *16) | PW _{IN(off)} | 0.7 | - | - | μs |
| PWM Input frequency | f _{РWМ} | - | - | 20 | kHz |
| Operating Junction Temperature | T _{j(ope)} | -30 | - | 150 | °C |

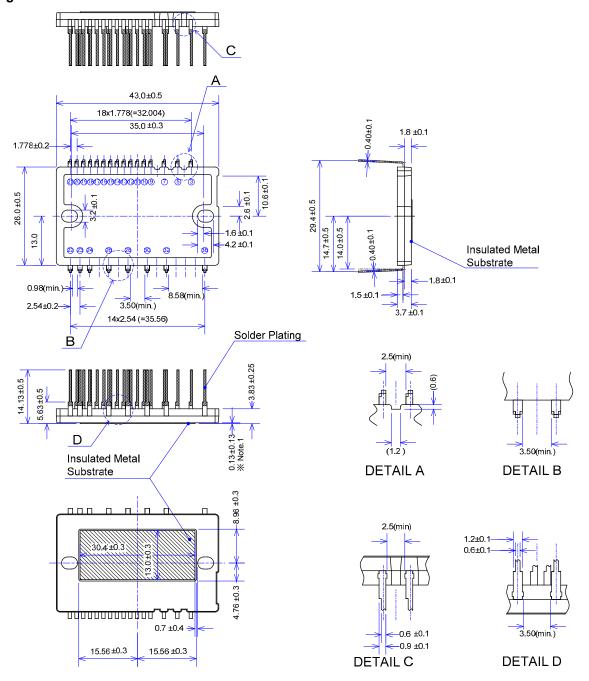
Note *14 : V_{DC}=300V, V_{CCH}=V_{CCL}=V_B(*)=15V, PF=0.8, Sinusoidal PWM, 3phase modulation, T_J≤150°C ,Tc≤100°C , f_{PWM}=5kHz, fo=200Hz, Ks=0.9

Note *15: In the pulse width of 0.5us, the loss of IGBT increases for the saturation operation.

To reduce the loss of IGBT, please enlarge the pulse width more than the switching time of IGBT.

Note *16: This IPM module might response according to input signal pulse even when the input signal pulse width is less than PW_{IN(on)} and PW_{IN(off)}.

■ Package outline dimensions



Unit: mm

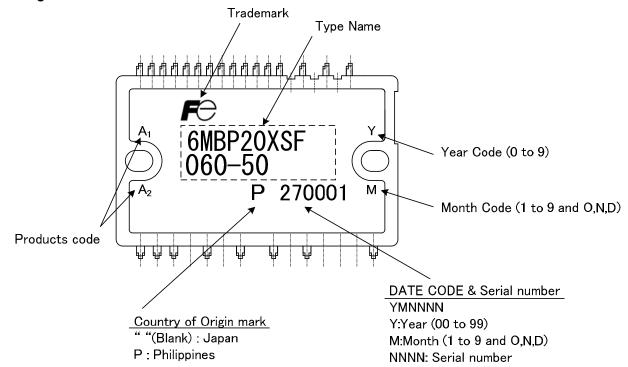
Note %1: The IMS(Insulated Metal Substrate) deliberately protruded from back surface of case. It is improved of thermal conductivity between IMS and heat-sink.

| Pin No. | Pin Name |
|---------|----------|
| 3 | VB(U) |
| 5 | VB(V) |
| 7 | VB(W) |
| 9 | IN(HU) |
| 10 | IN(HV) |
| 11 | IN(HW) |
| 12 | Vссн |
| 13 | COM |

| Pin No. | Pin Name |
|---------|------------------|
| 14 | IN(LU) |
| 15 | IN(LV) |
| 16 | IN(LW) |
| 17 | V _{CCL} |
| 18 | VFO |
| 19 | IS |
| 20 | COM |
| 21 | TEMP |
| | |

| Pin No. | Pin Name |
|---------|----------|
| 22 | N(W) |
| 23 | N(V) |
| 24 | N(U) |
| 26 | W |
| 28 | V |
| 30 | U |
| 32 | Р |
| 36 | NC |
| | |

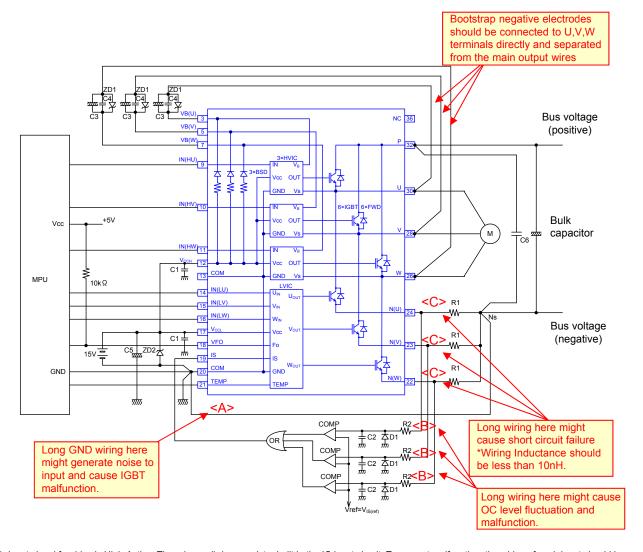
Marking



Note : Product code A_1 means current ratings , and "M" is marked. Product code A_2 means variations , and "F" is marked.

6MBP20XSF060-50

- An example of application circuit.
- Fig. shows an example of an application circuit.



- Note *1: Input signal for drive is High-Active. There is a pull-down resistor built in the IC input circuit. To prevent malfunction, the wiring of each input should be as short as possible. When using R-C coupling circuit, make sure the input signal level meet the turn-on and turn-off threshold voltage.
- Note *2: By the function of the HVIC, it is possible of the direct coupling to microprocessor (MPU) without any photo-coupler or pulse-transformer isolation.
- Note *3: VFO output is open drain type. It should be pulled up to the positive side of a 5V power supply by a resistor of about 10kW.
- Note *4: To prevent erroneous protection, the wiring of (A), (B), (C) should be as short as possible.
- Note *5: The time constant R2-C2 of the protection circuit should be selected approximately 1.5µs.
 - Over current (OC) shutdown time might vary due to the wiring pattern. Tight tolerance, temp-compensated type is recommended for R2, C2.
- Note *6: Please set the threshold voltage of the comparator reference input to be same as the IPM OC trip reference voltage V_{IS(ref)}.
- Note *7: Please use high speed type comparator and logic IC to detect OC condition quickly.
- Note *8: If negative voltage of R1 at the switching timing is applied, the schottky barrier diode D1 is recommended to be inserted parallel to R1.
- Note *9: All capacitors should be mounted as close to the terminals of the IPM as possible. (C1, C4: narrow temperature drift, higher frequency and DC bias characteristic ceramic type are recommended, and C3, C5: narrow temperature drift, higher frequency and electrolytic type.)
- Note *10: To prevent surge destruction, the wiring between the snubber capacitor and the P terminal ,Ns node should be as short as possible. Generally a 0.1μ to 0.22μF snubber capacitor (C6) between the P terminal and Ns node is recommended.
- Note *11: Two COM terminals (13 & 20 pin) are connected inside the IPM, it must be connected either one to the signal GND outside and leave another one open.
- Note *12: It is recommended to insert a zener-diode (22V) between each pair of control supply terminals to prevent surge destruction.
- Note *13: If signal GND is connected to power GND by broad pattern, it may cause malfunction by power GND fluctuation. It is recommended to connect signal GND and power GND at only a point.

Fig.1-2: The measurement position of heat sink flatness

Fig.1-1: The measurement position of temperature sensor.

Temperature sensor position

Approx. 7.0

Approx. 6.3

Heat sink side

To measurement position

SIDE VIEW

TOP VIEW

Fig.2-1: Switching waveforms

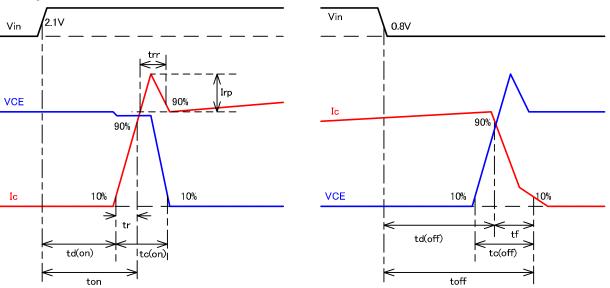
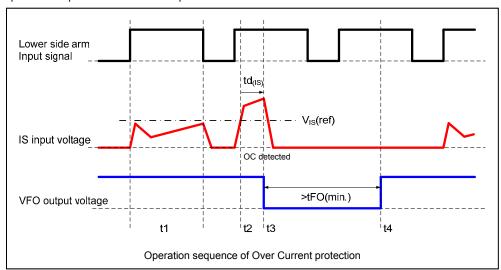
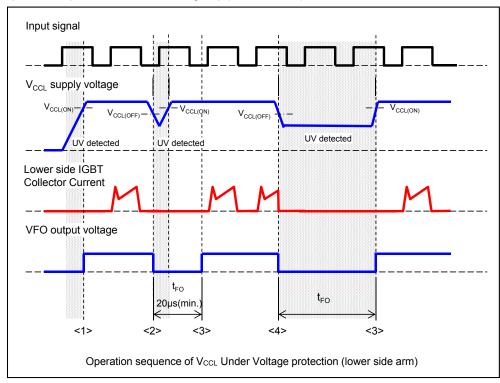


Fig.2-2: Operation sequence of Over current protection



- t1: IS input voltage does not exceed Vis(ref), while the collector current of the lower side IGBT is under the normal operation.
- t2: When IS input voltage exceeds Vis(ref), the OC is detected.
- t3: The fault output VFO is activated and all lower side IGBT shut down simultaneously after the over current protection delay time td(IS). Inherently there is dead time of LVIC in td(IS)
- t4: After the fault output pulse width tFO, the OC is reset. Then next input signal is activated.

Fig.2-3: Operation sequence of $V_{\text{\tiny CCL}}$ Under voltage trip (lower side arm)



When VccL is under 4V, UV and fault output are not activated.

- When VccL is under VccLoN, all lower side IGBTs are OFF state.

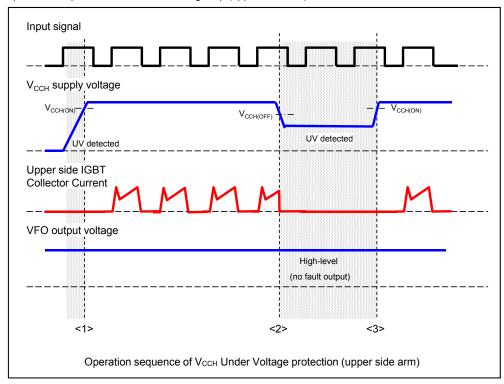
 After VccL rises VccLoN, the fault output VFO is released (high level).

 And the LVIC starts to operate, then next input is activated.

 42> The fault output VFO is activated when VccL falls below VccLoFF), and all lower side IGBT remains OFF state. When the voltage drop time is less than 20µs, the fault output pulse width is generated minimum 20µs and all lower side IGBTs are OFF state in spite of input signal condition during that time.
- <3> UV is reset after the when VccL exceeds VccL(ON) and the fault output VFO is reset simultaneously.
- And the LVIC starts to operate, then next input is activated.

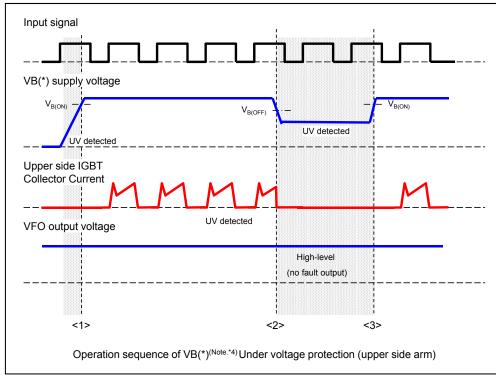
 <4> When the voltage drop time is more than t_{FO}, the fault output pulse width is generated and all lower side IGBTs are OFF state in spite of input signal condition during the same time.

Fig.2-4: Operation sequence of V_{CCH} Under voltage trip (upper side arm)



- <1> When V_{CCH} is under V_{CCH(ON)}, the upper side IGBT is OFF state.
 After V_{CCH} exceeds V_{CCH(ON)}, the HVIC starts to operate. Then next input is activated.
 The fault output VFO is constant (high level) not to depend on V_{CCH}.
- <2> After Vcch falls below VcchloFF, the upper side IGBT remains OFF state. But the fault output VFO keeps high level.
- <3> The HVIC starts to operate after UV is reset, then next input is activated.

Fig.2-5: Operation sequence of VB Under voltage trip (upper side arm)



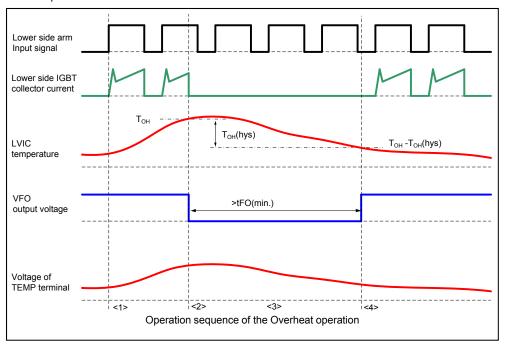
<1> When VB(*) is under $V_{B(ON)}$, the upper side IGBT is OFF state. After VB(*) exceeds $V_{B(ON)}$, the HVIC starts to operate. Then next input is activated. The fault output VFO is constant (high level) not to depend on VB(*). (Note*14)

Note *14: The fault output is not given HVIC bias conditions.

<2> After VB(*) falls below V_{B(OFF)}, the upper side IGBT remains OFF state. But the fault output VFO keeps high level.

<3> The HVIC starts to operate after UV is reset, then next input is activated.

Fig.2-6: Overheat protection



This function is applied to "6MBP**XSF060-50".

The IPM has the over-heating protection (OH) function by monitoring the LVIC temperature.

The Toh sensor position is shown in Fig.1-1.

- <1> The collector current of the lower side IGBT is under the normal operation while the LVIC temperature does not exceed ToH.
- <2> The IPM shutdown all lower side IGBTs while the LVIC temperature exceeds Toh.
- <3> The TEMP terminal continue to output the voltage which correspond to temperature of LVIC even if IPM is in OH condition.
- <4> The fault status is reset when the LVIC temperature drops below (ToH-ToH (hys)). The all lower side IGBTs restarted to normal operation.

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- Measurement equipment

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IGBT Modules

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