## LOW VOLTAGE AUDIO POWER AMPLIFIER

## - GENERAL DESCRIPTION

The NJM386 is a power amplifier designed for use in low voltage consumer applications. The gain is internally set to 20 to keep external part count low, but the addition of an external resistor and capacitor between pins 1 and 8 will increase the gain to any value up to 200 .

The inputs are ground reference while the output is automatically biased to one half the supply voltage. The quiescent power drain is only 24 milliwatts when operating from a 6 volt supply, making the NJM386 ideal for battery operation.

## - FEATURES

- Operating Voltage $(4 \mathrm{~V} \sim 12 \mathrm{~V})$
- Minimum External Components
- Low Operating Current ( 3 mA )
- Voltage Gain (20~200)
- Single Supply Operation
- Self-centering of Output Offset Voltage
- Package Outline DIP8, SIP8, DMP8
- Bipolar Technology
- A PPLICATIONS
- AM-FM radio amplifiers
- Portable tape player amplifiers
- Intercoms
- TV sound systems
- Line drivers
- Ultrasconic drivers
- Small servo drivers
- Power converters
- PIN CONFIGURATION

- PACKAGE OUTLINE


HSM386M


NJM386L

PIN CONNECTION

1. GAIN
2. -INPUT
3. +INPUT
4. GND
5. OUTPUT
6. $\mathrm{V}^{+}$
7. BYPASS
8. GAIN

## EQUIVALENT CIRCUIT



- ABSOLUTE MAXIMUM RATINGS
( $\mathrm{ta}=25^{\circ} \mathrm{C}$ )

| PARAMETER | SYMBOL | RATINGS | UNIT |
| :--- | :---: | :---: | :---: |
| Supply Voltage | $\mathrm{V}^{+}$ | 15 | V |
| Power Dissipation | PD | (DIP8) 700 | mW |
|  |  | $($ SIP8 $) 800$ | mW |
|  |  | (DMP8) 300 | mW |
| Input Voltage Range | $\mathrm{V}_{\text {IN }}$ | $\pm 0.4$ | V |
| Operating Temperature Range | $\mathrm{T}_{\text {opr }}$ | $-40 \sim+85$ | ${ }^{\circ} \mathrm{C}$ |
| Storage Temperature Range | $\mathrm{T}_{\mathrm{stg}}$ | $-40 \sim+125$ | ${ }^{\circ} \mathrm{C}$ |

- ELECTRICAL CHARACTERISTICS
$\left(\mathrm{Ta}=25^{\circ} \mathrm{C}\right)$

| PARAMETER | SYMBOL | TEST CONDITION | MIN. | TYP. | MAX. | UNIT: |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Operating Voltage ${ }^{\text {' }}$ | $V^{+}$ |  | 4 | - | 12 | V |
| Operating Current | 1 cc | $\mathrm{V}^{+}=6 \mathrm{~V}, \mathrm{~V}_{\text {IN }}=0$ | - | 3 | 8 | mA |
| Output Power (note 2) | $\mathrm{P}_{0}$ | $V^{+}=6 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=8 \Omega, \mathrm{THD}=10 \%$ | 250 | 325 | - | mW |
|  |  | $\mathrm{V}^{+}=9 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=16 \Omega, \mathrm{THD}=10 \%$ | - | 500 | - | mW |
| Voltage Gain | Av | $\mathrm{V}^{+}=6 \mathrm{f}, \mathrm{f}=1 \mathrm{kHz}$ | 24 | 26 | 28 | dB |
|  |  | $10 \mu \mathrm{~F}$ from Pin 1 to 8 | 43 | 46 | 49 | dB |
| Bandwidth | BW | $\mathrm{V}^{+}=6 \mathrm{~V}$, Pins 1 and 8 Open |  | 300 | - | kHz |
| Total Harmonic Distortion | THD | $\begin{aligned} & V^{+}=6 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=8 \Omega, \text { Pout }=125 \mathrm{~mW} \\ & \Gamma=1 \mathrm{kHz}, \text { Pins } 1 \text { and } 8 \text { open } \end{aligned}$ | - | 0.2 | - | \% |
| Power Supply Rejection Ratio | SVR | $\mathrm{V}^{+}=6 \mathrm{~V}, \mathrm{f}=1 \mathrm{kHz}, \mathrm{C}_{\text {BYPASS }}=10 \mu \mathrm{~F}$ <br> Pins 1 and 8 Open | - | 50 | - | dB |
| Input Resistance | $\mathrm{R}_{\mathrm{IN}}$ |  | - | 50 | - | $k \Omega$ |
| Input Bias Current | $\mathrm{I}_{13}$ | $V^{+}=6 \mathrm{~V}$, Pins 2 and 3 Open | - | 250 | - | nA |

## - TYPICAL APPLICATION



## POWER DISSIPATION VS. AMBIENT TEMPERATURE



## - NOTICE WHEN APPLICATION

## - Prevention of Oscillation

It is recommended to insert capacitors at around the supply source and the GND pins with the value of $0.1 \mu \mathrm{~F}$ and more than $100 \mu \mathrm{~F}$ which are featuring higher frequency efficiency.

When statt of oscillation accordingly to the load condition, it is recommendable to insert the resistor of $10 \Omega$ and the capacitor of $0.047 \mu \mathrm{~F}$ between the output and the GND pins.

- How to use the Input Resistor (TYP. $50 \mathrm{k} \Omega$ )

The input resistors have much deviation in value generally, so that it is recommended not to use them as the constant of the circuit. The countermesure to be recommended si to apply the resistor of higher in value, which is so higher to be able to ignore the input deviation( $3 \mathrm{k} \Omega$ approximately) in parallel application.


## - Maintenance of Output Offset Voltage

By making connection of both input pins with low value resistors (below $10 \mathrm{~K} \Omega$ approximately) to GND, the output offset voltage is automatically set in the medium range value of the supply source. However, the DC Gain of NJM386 is approximately at 20 times in value, so that when keeping one side input pin open, and the other side to GND on DC condition. The voltage drop caused by input resistor $X$ input bias current, that is, (input resistor $X$ input bias current) $X 20$ times voltage is to be added to the output offset voltage, and that the medium range output voltage is to be sheared, which in the result, no distortion output oscillation range shall be decreased.

In regard to dealing with the input pin, it is recommendable to put the input pin into the GND at first, and the other side of signal input pin, to be connected into GND with the resistor of less than about $10 \mathrm{~K} \Omega$ on DC condition.

## - Concerning Cross-Over Distortion

NJM386 in application, the cross-over distortion is to be generated in the high band operation.
The countermeasure for that, it is recommendable to have it replaced with NJM386B (But, be carful in prevention of oscillation). And for prevention of the cross-over distortion, it is recommendable to apply NJM2072, NJM2073.

- The Application Purpose and Recommended Value of the External parts.

| EXTERNAL PARTS | APPLICATION PURPOSE | RECOMMENDED VALUE | REMARKS |
| :---: | :---: | :---: | :---: |
| Rs | Current like nois reduction $V_{0 Q}$ stabilization | Below $10 \mathrm{~K} \Omega$ | The noise becomes high when the input pin opend. |
| $C_{\text {IN }}$ | $\mathrm{V}_{0 \mathrm{Q}}$ stabilization | $1 \mu \mathrm{~F}$ | It is not required in case when there is no $D C$ offset in the input signal. |
| CPI | $\mathrm{V}^{+}$stabilization | こCCup | It can be decreased in value when the output impedance source is low. |
| $\mathrm{Cp}_{2}$ | Oscillation prevention | $0.1 \mu \mathrm{~F}$ | Insert near around the supply source and GND pins. |
| $\mathrm{Cb}_{\mathrm{B}}$ | Ripple rejection to Voby way of $\mathrm{V}^{+}$ | $47 \mu \mathrm{~F}$ | It is not required when the $\mathrm{V}^{+}$is stabilized. |
| Co | Oscillation prevention | $0.047 \mu \mathrm{~F}$ | To be decided in value according to load condition. |
| $\mathrm{R}_{0}$ | Oscillation prevention | $10 \Omega$ | To be decided in value according to load condition. |
| Ccup | Output DC Decoupling | $\begin{aligned} & 220 \mu \mathrm{~F} \text { when } \\ & \mathrm{R}_{\mathrm{L}}=8 \Omega \end{aligned}$ | Low band cutoff frequency ( $f_{\mathrm{L}}$ ) shall be decided by $\mathrm{C}_{\mathrm{Cu}} \mathrm{R}_{\mathrm{L}}$. When $C_{\text {Cup }}$ is less in value, $f_{L}$ is to be increased. |



## - MUTING CIRCUIT EXAMPLE

(I) The way how to apply DC voltage to -INPUT pin.


According to this method, when applicating DC voltage, Vmute to -INPUT PIN, the output voltage $V_{0}$ at voltage gain Av will be,
$\mathrm{V}_{0}=\mathrm{V}^{+} / 2-\mathrm{Vmute}^{*} \mathrm{~A}_{\mathrm{V}}$
It is the way that the muting shall be proceeded by keeping $V_{o}$ saturating at the GND side. Now, the output is saturated, so that there is no leakage of muting. However, when the peak value of signal input is increased higher than about the value of $1 / 4$ Vmute, the leakage of muting shall be started.
(2) The way, how to connect gain. No. 8 PIN to GND


It is the way, originally that the pin which is to be used for adjusting the gain of NLM386, but to have it applied in connecting to GND side, and by doing so, to stop the earely stage motion, but keeping on for muting operation. The earely stage motion shall be stopped, therefore, the precise muting shall be proceeded with less leakage on operation.
(3) The way how to proceed casting the BY PASS pin on $\mathrm{V}^{+}$side


By this way, the bias circuit within IC, to be stopped and then, further for stopping motion of driver level, and at the output level. However, the input level alone is operating, so that a slight leakage of signal to the output pin through inside resistor to be occured. The leakage level is to be inverse proportion to load, therefore, it is necessary to check accordingly through the load condition.
(Note) Improper Muting Circuit
Never to apply with the Muting Circuit, because of the fact that, there are cases when the muting does not operate depending on IC to be used.
The way how to connect the BY PASS PIN to GND.


## - APPLICATION CIRCUIT EXAMPLE

Amplifier 1


Amplifier 2


Low Distortion Power Wienbridge Oscillator


Square Wave Oscillator


## - WIDE RANGE APPLICATION

NJM386 is a small output power amplifier with minimum external parts, and also the gain of which is fixed, yet it can be made changeable in value, too.

## GAIN CONTROL

To make the NJM386 a more versatile amplifier, two pins (I and 8) are provieded the gain contorol. With pins I and 8 open the $1.35 \mathrm{k} \Omega$ resistor sets the gain at $20(26 \mathrm{~dB})$. If a capacitor is put from pin 1 to 8 , bypassing the $1.35 \mathrm{k} \Omega$ resistor, the gain will go up to $200(46 d B)$. If a resistor is placed in series with the capacitor. the gain can be set to any value from 20 to 200. Gain contorol can also be done by capacitively coupling a resistor (or FET) from pin 1 to ground.

Additional external components can be placed in parallel with the internal feedback resistors to tailor the gain and frequency response for individual appapplications. For example, we can compensate poor speaker bass response by frequency shaping the feedback path. This is done with a series $R C$ from pin 1 to 5 (paralleling the internal $I 5 k \Omega$ resistor). For $6 d B$ effective bass boost: $R \cong 15 \mathrm{k} \Omega$, the lowest value for good stable operation is $R_{\text {miN }}=10 \mathrm{k} \Omega$ if pin 8 is open. If pins 1 and 8 are bypassed then $R$ as low as $2 k \Omega$ can be used. This restriciton is because the amplifier is only compensated for closed-loop gains greater than 9 .

- TYPICAL CHARACTERISTICS

Operating Current vs. Operating Voltage


Voltage Gain vs. Frequency


Total Harmonic Distortion vs. Frequency $\left(\mathrm{V}^{+}=6 \mathrm{~V}, \mathrm{R}_{1}=8 \Omega, \mathrm{P}_{0}=125 \mathrm{~mW}, \mathrm{Av}=26 \mathrm{~dB}, \mathrm{Ta}=25^{\circ} \mathrm{C}\right)$


## Maximum Output Voltage Swing

vs. Operating; Voltage


## Power Supply Rejection Ratio

vs. Frequency
$\left(\mathrm{V}^{+}=6 \mathrm{~V}, \mathrm{Av}=26 \mathrm{~dB}, \mathrm{Ta}=25^{\circ} \mathrm{C}\right)$


Total Harmonic Distortion
vs. Output Power


- TYPICAL CHARACTERISTICS

Power Dissipation vs. Output Power


Power Dissipation vs. Output Power


Frequency Response with Bass Boost

(Typical Application "Amplifier 1")

Power Dissipation vs. Output Power


Power Dissipation vs. Output Power


## - TYPICAL CHARACTERISTICS



Voltage Gain vs. Temperature


## MEMO

[CAUTION]
The specifications on this databook are only given for information, without any guarantee as regards either mistakes or omissions. The application circuits in this databook are

## X-ON Electronics

Largest Supplier of Electrical and Electronic Components
Click to view similar products for Audio Amplifiers category:
Click to view products by NJR manufacturer:
Other Similar products are found below :
LV47002P-E MP7747DQ-LF-P AZ386MTR-E1 NCP2811AFCT1G NCP2890AFCT2G NJM8068RB1-TE1 NJW1194V-TE1 LA4282-E LA4814JA-AE LC706200CM SSM2377ACBZ-R7 FDA2100LV TDA2541 TDA7385H TDA7391LV TDA7575BPDTR TDA7718NTR IS31AP2121-LQLS1 IS31AP4915A-QFLS2-TR LA74309FA-BH 421067X 480263C NCP2820FCT2G STPA001 TDA1515AQ TDA1520B TDA1591T TDA2051H TDA4850 TDA7391PDUTR TDA7563BH TDA7718B LA4425F-E LA4742-E TDA7391PDU TDA7491MV13TR TDA749213TR TDA7563AH TDA7850H STK433-070GN-E E-TDA7391PDTR SSM2529ACBZ-R7 SSM2518CBZR7 MAX9890BEBL+T MAX98303EWE+T MAX98358EWL+ MAX98304DEWL+T MAX97220DETE+T TS4962MEIJT TS4990EIJT

