

Features

■ LOW OFFSET VOLTAGE: 10 µV (Max)

■ ZERO DRIFT: 0.008 µV/°C

0.1Hz to 10Hz Noise: 1.1 μV_{PP}

■ Low Supply Current: 34 µA per Amplifier

■ Bandwidth: 350 kHz ■ Slew Rate: 0.16 V/µs

■ High Gain, 130 dB High CMRR and PSRR

■ Rail-to-rail Input and Output Swing

■ -40°C to 125°C Operation Range

■ Small Packages: SC70 and SOT23 (TP5531)

Applications

- Transducer Amplifier
- Bidirectional Current Sense
- DC Offset Correction
- Temperature Measurement
- Remote Located Sensors
- Battery-Powered Instruments
- Electronic Weigh Scales

Description

The 3PEAK TP5531/2/4 low-power chopper stabilized operational amplifiers provide input offset voltage correction for very low offset and offset drift over time and temperature. The devices operate with a single supply voltage as low as 1.8V, while drawing 34µA per amplifier of quiescent current with a gain bandwidth product of 350kHz. They are unity gain stable, have no 1/f noise, have good Power Supply Rejection Ratio (PSRR) and Common Mode Rejection Ratio (CMRR), and feature rail-to-rail input and output swing.

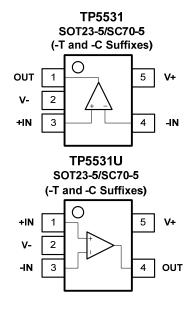
The devices were designed using an advanced CMOS process. The TP5531 (single version) is available in SC70-5, SOT23-5 and SO-8 packages. The TP5532 (dual version) is offered in MSOP-8 and SO-8 package. The TP5534 (quad version) is available in TSSOP-14 and SOIC-14 package. All versions are specified for operation from -40°C to 125°C.

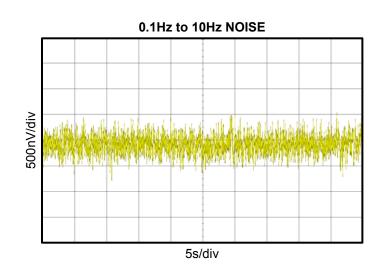
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Related Zero-Drift Op-amps

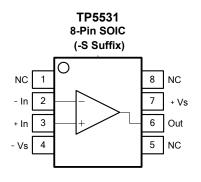
| Vos (Max.) | 10 μV | 5 μV | 5 μV |
|-------------------------|-----------|-----------|-----------|
| GBWP | 350 kHz | 1.5 MHz | 3.5 MHz |
| Supply Current | 34 μΑ | 220 μΑ | 500 μΑ |
| e _N at 1 kHz | 55 nV/√Hz | 25 nV/√Hz | 15 nV/√Hz |
| Single | TP5531 | TP5541 | TP5551 |
| Dual | TP5532 | TP5542 | TP5552 |
| Quad | TP5534 | TP5544 | TP5554 |

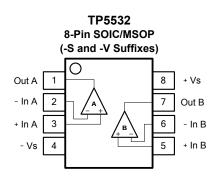
Pin Configuration (Top View)

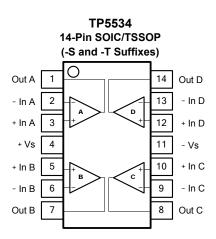




Pin Configuration (Top View, continued)







Order Information

| Model Name | Order Number | Package | Transport Media, Quantity | Marking Information |
|------------|--------------|-----------------|---------------------------|------------------------|
| | TP5531-TR | SOT23-5 | Tape and Reel, 3,000 | E31T |
| TP5531 | TP5531-CR | SC70-5 (SOT353) | Tape and Reel, 3,000 | E31C |
| | TP5531-SR | SOIC-8 | Tape and Reel, 4,000 | E31S |
| TP5531U | TP5531U-TR | SOT23-5 | Tape and Reel, 3,000 | E31U |
| | TP5531U-CR | SC70-5 | Tape and Reel, 3,000 | E31V |
| TP5532 | TP5532-SR | SOIC-8 | Tape and Reel, 4,000 | E32S |
| 173332 | TP5532-VR | MSOP-8 | Tape and Reel, 3,000 | E32V |
| TP5534 | TP5534-SR | SOIC-14 | Tape and Reel, 2,500 | E34S |
| 11 3334 | TP5534-TR | TSSOP-14 | Tape and Reel, 3,000 | E34T |

Absolute Maximum Ratings Note 1

| Supply Voltage: | 6V |
|---|-------------------------|
| Input Voltage: | $V^ 0.2$ to $V^+ + 0.2$ |
| Input Current: +IN, -IN Note 2 | ±20mA |
| Output Current: OUT | ±60mA |
| Output Short-Circuit Duration Note | 3Indefinite |

| Current at Supply Pins | +50m∆ |
|--------------------------------------|----------|
| Operating Temperature Range40°C t | |
| | |
| Maximum Junction Temperature | |
| Storage Temperature Range –65°C t | to 150°C |
| Lead Temperature (Soldering, 10 sec) | 260°C |

Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

Note 2: The inputs are protected by ESD protection diodes to each power supply. If the input extends more than 500mV beyond the power supply, the input current should be limited to less than 10mA.

Note 3: A heat sink may be required to keep the junction temperature below the absolute maximum. This depends on the power supply voltage and how many amplifiers are shorted. Thermal resistance varies with the amount of PC board metal connected to the package. The specified values are for short traces connected to the leads.

ESD, Electrostatic Discharge Protection

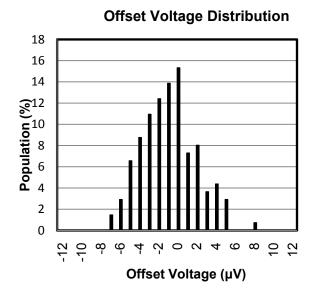
| Symbol | Parameter | Condition | Minimum Level | Unit |
|--------|--------------------------|----------------------------|---------------|------|
| HBM | Human Body Model ESD | MIL-STD-883H Method 3015.8 | 7 | kV |
| CDM | Charged Device Model ESD | JEDEC-EIA/JESD22-C101E | 2 | kV |

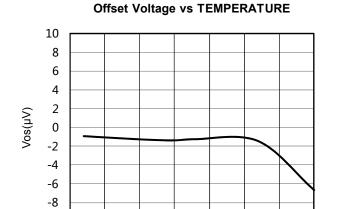
Electrical Characteristics

The boldface denotes the specifications which apply over the full operating temperature range, T_A = -40°C to +125°C. At T_A = 27°C, V_{DD} = 5V, R_L = 10k Ω , V_{CM} = $V_{DD}/2$, unless otherwise noted.

| SYMBOL | PARAMETER | CONDITIONS | MIN | TYP | MAX | UNITS |
|-----------------------|--------------------------------------|--------------------------------|------------|-------|-----------------------|-----------------|
| V_{DD} | Supply Voltage Range | | 1.8 | | 5.5 | V |
| ΙQ | Quiescent current per amplifier | I _O = 0 | | 34 | 40 | μA |
| | Over temperature | | | | 55 | μΑ |
| Vos | Input Offset Voltage | Input grounded, unity gain. | | 2 | 10 | μV |
| $dV_{\text{OS}}\!/dT$ | vs temperature | | | 0.008 | 0.05 | μV/°C |
| PSRR | vs power supply | Vs = +1.8V to +5.5V | | 0.5 | | μV/V |
| Vn | input voltage noise, f=0.01Hz to 1Hz | | | 0.4 | | μV_{pp} |
| | input voltage noise, f=0.1Hz to 10Hz | | | 1.1 | | μV_{pp} |
| en | Input voltage noise density, f=1kHz | | | 55 | | nV/√Hz |
| Cin | Input capacitor, Differential | | | 3 | | pF |
| | Input capacitor, Common-Mode | | | 2 | | pF |
| I_B | Input Bias Current Over temperature | | | ±50 | ±200 800 | рА рА |
| los | Input offset current | | | ±100 | ±400 | pА |
| V _{CM} | Common-mode voltage range | | (V-) - 0.1 | | (V ₊)+0.1 | V |
| CMRR | Common-mode rejection ratio | | 110 | 130 | | dB |
| V_{0} | Output Voltage Swing from rail | R _L =10kΩ | | 5 | 10 | mV |
| | Over temperature | R _L =10kΩ | | | 10 | mV |
| Isc | Short-circuit current | | | | ±60 | mA |
| CL | Maximum Capacitive Load | | | 1,000 | | pF |
| GBWP | Unity Gain Bandwidth | C _L =100pF | | 350 | | kHz |
| SR | Slew rate | G=+1, CL=100pF | | 0.16 | | V/µs |
| tor | Overload recovery time | G=-10 | | 60 | | μs |
| ts | Settling time to 0.01% | C _L =100pF | | 40 | | μs |
| Avol | Open-Loop Voltage Gain | $(V-)+100mV< V_0< (V+)-100mV,$ | 100 | 120 | | dB |
| 7.1702 | open 200p remage came | $R_L = 100k\Omega$ | 100 | | | 4.2 |
| | | SC70-5 (SOT353) | | 250 | | |
| | | SOT23-5 | | 200 | | |
| θ.JA | Thermal Resistance Junction to | MSOP-8 | | 210 | | °C/W |
| UJA | Ambient | SOIC-8 | | 158 | | O/VV |
| | | SOIC-14 | | 83 | | |
| | | TSSOP-14 | | 100 | | |

Typical Performance Characteristics





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Temperature(°C)

50

75

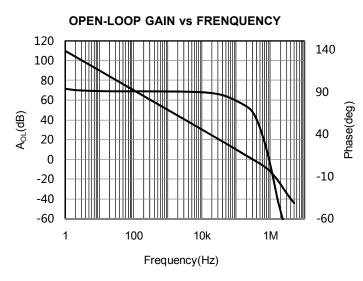
100

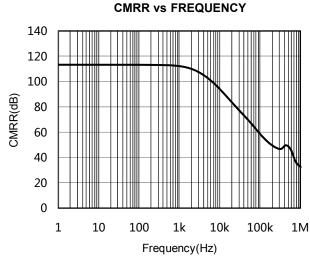
125

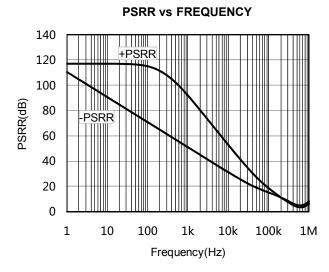
-10

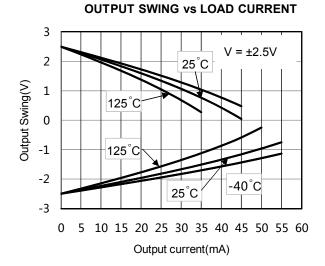
-50

-25

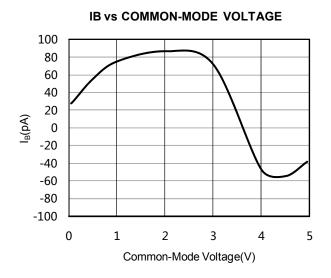


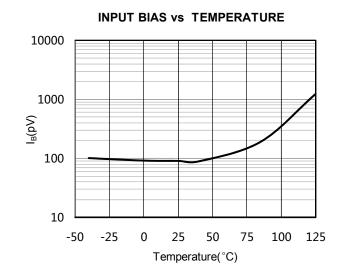




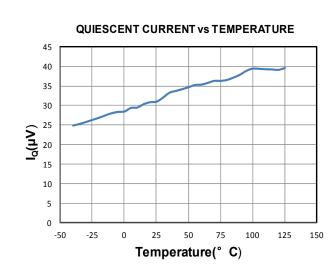


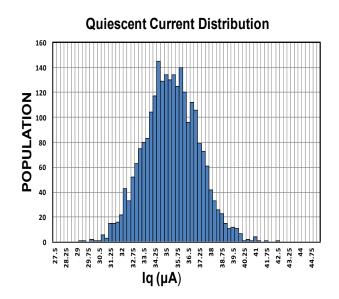
Typical Performance Characteristics(continue)

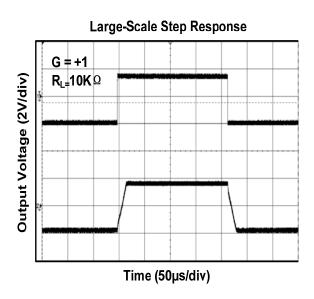




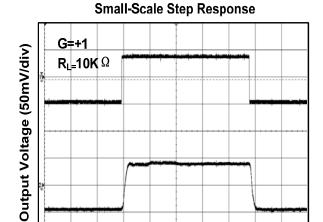
Vosvs COMMON-MODE VOLTAGE 2 1.5 1 0.5 Vos(µV) 0 -0.5 -1 -1.5 -2 0 1 2 3 5 Common-Mode Voltage(V)



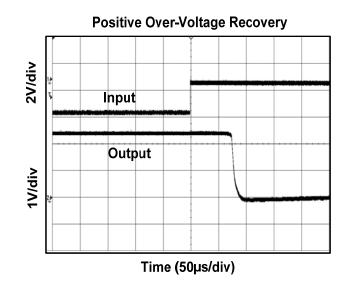


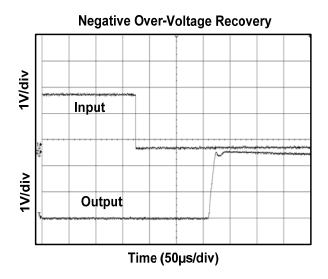


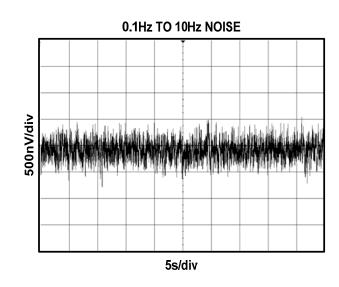
Typical Performance Characteristics (continue)

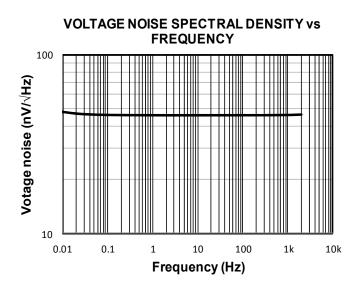


Time (5µs/div)





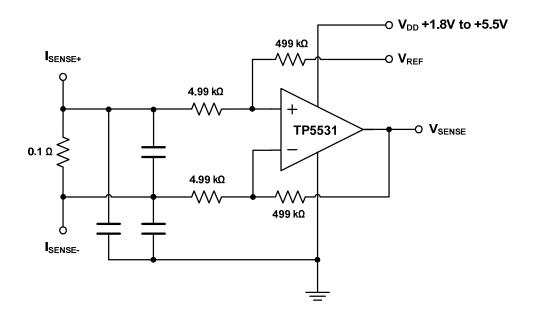




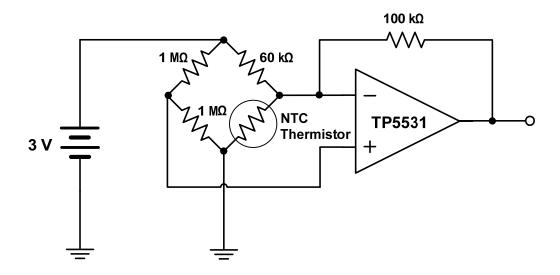
Typical Applications

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Bi-Directional Current Sense Amplifier



Thermistor Measurement



Pin Functions

-IN: Inverting Input of the Amplifier.

+IN: Non-Inverting Input of Amplifier.

OUT: Amplifier Output. The voltage range extends to within mV of each supply rail.

V+ or +V_s: Positive Power Supply. Typically the voltage is from 1.8V to 5.5V. Split supplies are possible as long as the voltage between V+ and V- is between 1.8V and 5.5V. A bypass capacitor of 0.1µF as close to the part as

possible should be used between power supply pins or between supply pins and ground.

V- or -V_s: Negative Power Supply. It is normally tied to ground. It can also be tied to a voltage other than ground as long as the voltage between V⁺ and V⁻ is from 1.8V to 5.5V. If it is not connected to ground, bypass it with a capacitor of 0.1μF as close to the part as possible.

Operation

The TP553x series op amps are zero drift, rail-to-rail operation amplifiers that can be run from a single-supply voltage. They use an auto-calibration technique with a time-continuous 350 kHz op amp in the signal path while consuming only 34 μ A of supply current per channel. This amplifier is zero-corrected with an 120 kHz clock. Upon power-up, the amplifier requires approximately 100 μ s to achieve specified V_{OS} accuracy. This design has no aliasing or flicker noise.

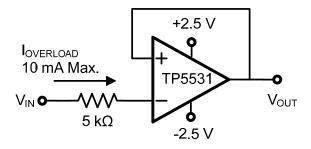
Applications Information

Rail-To-Rail Input And Output

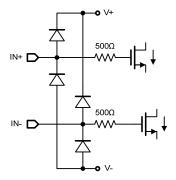
The TP553x series op amps feature rail-to-rail input and output with a supply voltage from 1.8V to 5.5 V. This allows the amplifier inputs to have a wide common mode range (50mV beyond supply rails) while maintaining high CMRR (130dB) and maximizes the signal to noise ratio of the amplifier by having the V_{OH} and V_{OL} levels be at the V+ and V-rails, respectively.

Input Protection

The TP553x series op amps have internal ESD protection diodes that are connect between the inputs and supply rail. When either input exceeds one of the supply rails by more than 300mV, the ESD diodes become forward biased and large amounts of current begin to flow through them. Without current limiting, this excessive fault current causes permanent damage to the device. Thus an external series resistor must be used to ensure the input currents never exceed 10mA.



Current-limiting resistor required if input voltage exceeds supply rails by >0.5V.



INPUT ESD DIODE CURRENT LIMITING-UNITY GAIN

Low Input Referred Noise

Flicker noise, as known as 1/f noise, is inherent in semiconductor devices and increases as frequency decreases. So at lower frequencies, flicker noise dominates, causing higher degrees of error for sub-Hertz frequencies or dc precision application.

The TP553x series amplifiers are chopper stabilized amplifiers, the flicker noise is reduced greatly because of this technique. This reduction in 1/f noise allows the TP553x to have much lower noise at dc and low frequency compared to standard low noise amplifier.

Residual Voltage Ripple

The chopping technique can be used in amplifier design due to the internal notch filter. Although the chopping related voltage ripple is suppressed, higher noise spectrum exists at the chopping frequency and its harmonics due to residual ripple.

So if the frequency of input signal is nearby the chopping frequency, the signal maybe interfered by the residue ripple. To further suppress the noise at the chopping frequency, it is recommended that a post filter be placed at the output of the amplifier.

Broad Band and External Resistor Noise Considerations

The total broadband noise output from any amplifier is primarily a function of three types of noise: input voltage noise from the amplifier, input current noise from the amplifier, and thermal (Johnson) noise from the external resistors used around the amplifier. These noise sources are not correlated with each other and their combined noise can be summed in a root sum squared manner. The full equation is given as:

$$e_n total = [e_n^2 + 4kTR_s + (i_n \times R_s)^2]^{1/2}$$

Where:

e_n= the input voltage noise density of the amplifier.

 i_0 = the input current noise of the amplifier.

 $R_{\rm S}$ = source resistance connected to the noninverting terminal.

k= Boltzmann' s constant (1.38x10⁻²³J/K). T= ambient temperature in Kelvin (K).

The total equivalent rms noise over a specific bandwidth is expressed as:

$$e_{n,rms} = e_n \ total \times \sqrt{BW}$$

The input voltage noise density (en) of the TP553x is 55 nV/ $\sqrt{\text{Hz}}$, and the input current noise can be neglected. When the source resistance is 190 k Ω , the voltage noise contribution from the source resistor and the amplifier are equal. With source resistance greater than 190 k Ω , the overall noise of the system is dominated by the Johnson noise of the resistor itself.

High Source Impedance Application

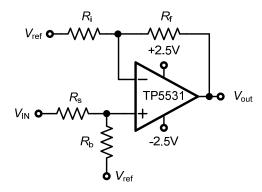
The TP553x series op amps use switches at the chopper amplifier input, the input signal is chopped at 125 kHz to reduce input offset voltage down to $10\mu V$. The dynamic behavior of these switches induces a charge injection current to the input terminals of the amplifier. The charge injection current has a DC path to ground through the resistances seen at the input terminals of the amplifier. Higher input impedance causes an apparent shift in the input bias current of the amplifier.

Because the chopper amplifier has charge injection currents at each terminal, the input offset current will be larger than standard amplifiers. The I_{OS} of TP553x are 150pA under the typical condition. So the input impedance should be

balanced across each input. The input impedance of the amplifier should be matched between the IN+ and IN-terminals to minimize total input offset current. Input offset currents show up as an additional output offset voltage, as shown in the following equation:

$$v_{os,total} = v_{os} - R_f \times I_{os}$$

For a gain configure using $1M\Omega$ feedback resistor, a 150pA total input offset current will have an additional output offset voltage of 0.15mV. By keeping the input impedance low and balanced across the amplifier inputs, the input offset current effect will be suppressed efficiently.



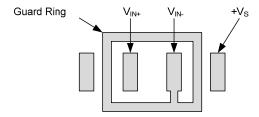
Circuit Implication for reducing Input offset current effect

PCB Surface Leakage

In applications where low input bias current is critical, Printed Circuit Board (PCB) surface leakage effects need to be considered. Surface leakage is caused by humidity, dust or other contamination on the board. It is recommended to use multi-layer PCB layout and route the OPA's -IN and +IN signal under the PCB surface.

The effective way to reduce surface leakage is to use a guard ring around sensitive pins (or traces). The guard ring is biased at the same voltage as the sensitive pin. An example of this type of layout is shown in Figure 2 for Inverting Gain application.

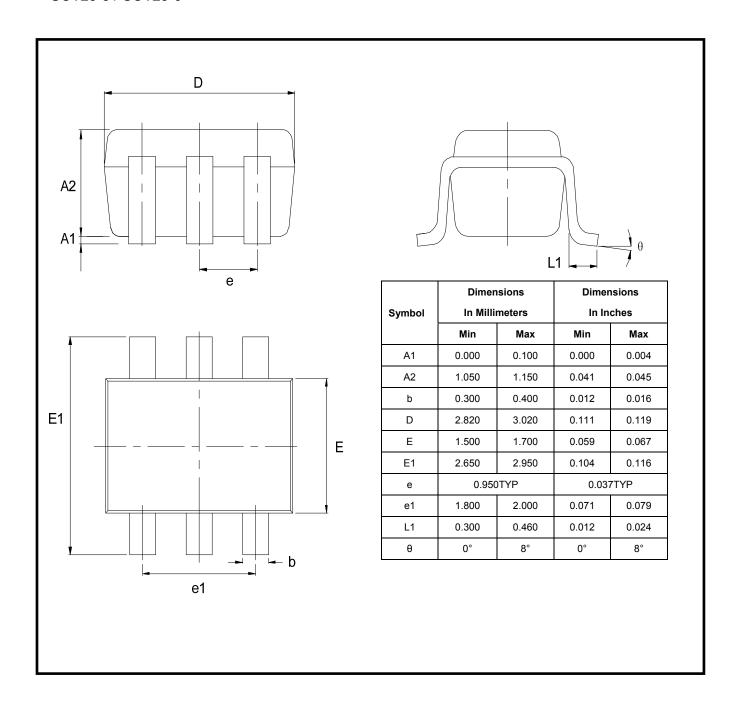
- 1. For Non-Inverting Gain and Unity-Gain Buffer:
 - a) Connect the non-inverting pin (V_{IN}+) to the input with a wire that does not touch the PCB surface.
 - b) Connect the guard ring to the inverting input pin (V_{IN}-). This biases the guard ring to the Common Mode input voltage.
- 2. For Inverting Gain and Trans-impedance Gain Amplifiers (convert current to voltage, such as photo detectors):
 - a) Connect the guard ring to the non-inverting input pin $(V_{IN}+)$. This biases the guard ring to the same reference voltage as the op-amp (e.g., $V_{DD}/2$ or ground).
 - b) Connect the inverting pin (V_{IN}-) to the input with a wire that does not touch the PCB surface.



The Layout of Guard Ring

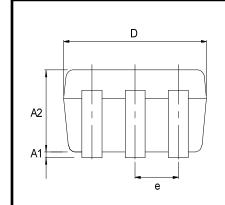
Package Outline Dimensions

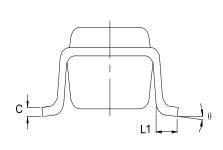
SOT23-5 / SOT23-6

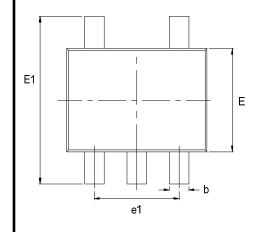


Package Outline Dimensions

SC-70-5 (SOT353)



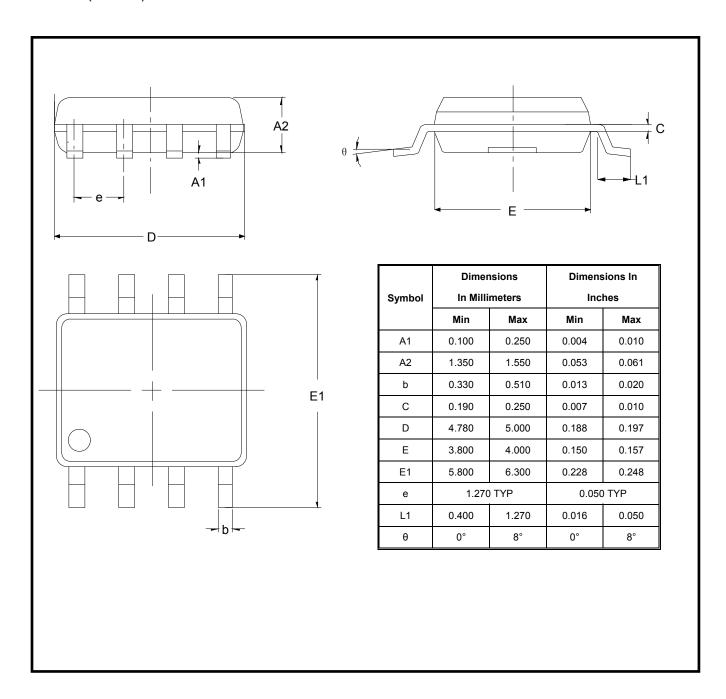




| | Dimensions | | Dimensions In | | |
|--------|----------------|-------|---------------|-------|--|
| Symbol | In Millimeters | | Inches | | |
| | Min | Max | Min | Max | |
| A1 | 0.000 | 0.100 | 0.000 | 0.004 | |
| A2 | 0.900 | 1.000 | 0.035 | 0.039 | |
| b | 0.150 | 0.350 | 0.006 | 0.014 | |
| С | 0.080 | 0.150 | 0.003 | 0.006 | |
| D | 2.000 | 2.200 | 0.079 | 0.087 | |
| E | 1.150 | 1.350 | 0.045 | 0.053 | |
| E1 | 2.150 | 2.450 | 0.085 | 0.096 | |
| е | 0.650TYP | | 0.026TYP | | |
| e1 | 1.200 | 1.400 | 0.047 | 0.055 | |
| L1 | 0.260 | 0.460 | 0.010 | 0.018 | |
| θ | 0° | 8° | 0° | 8° | |

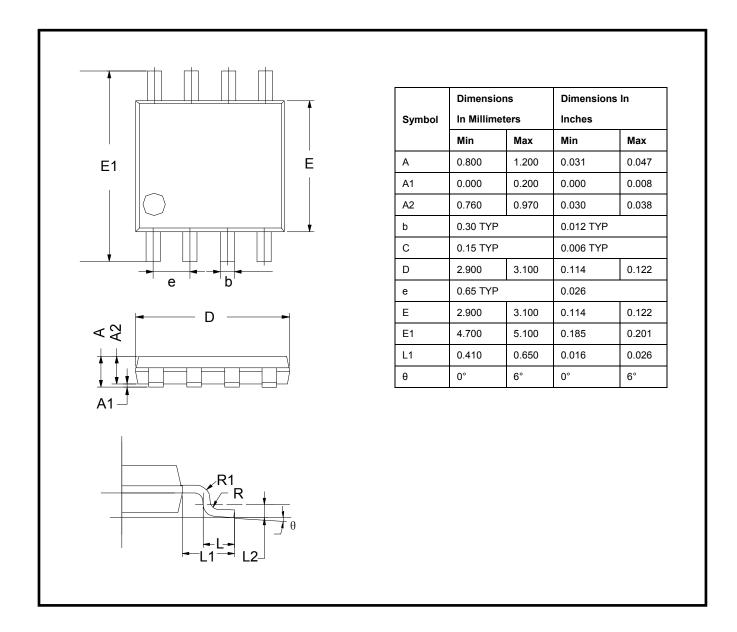
Package Outline Dimensions

SOP-8 (SOIC-8)



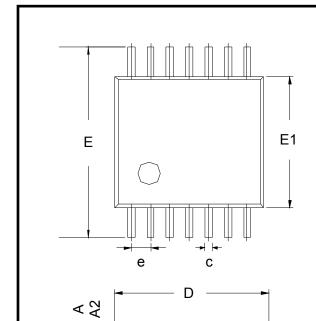
1.8V, 34µA, RRIO, Zero Drift Op-amps Package Outline Dimensions

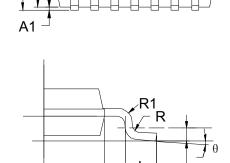
MSOP-8



Package Outline Dimensions

TSSOP-14

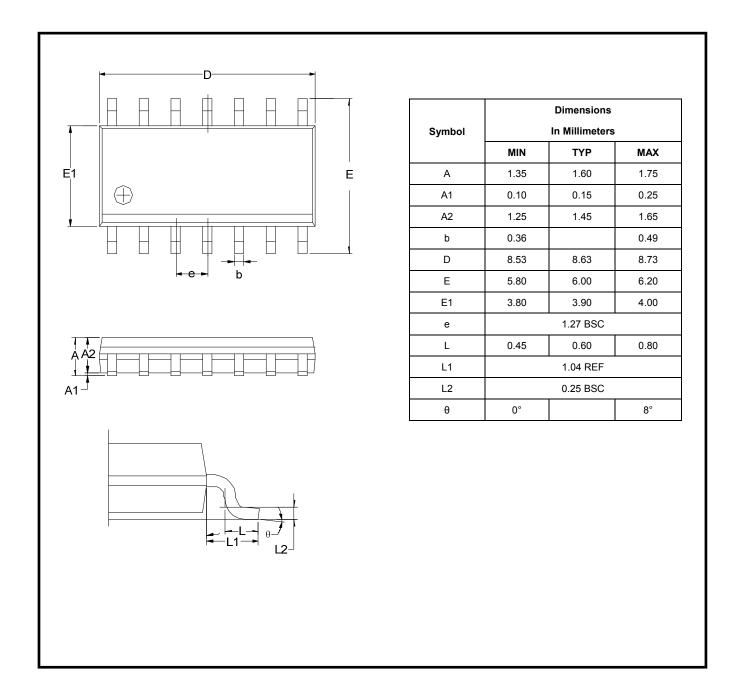




| | Dimensions In Millimeters | | | |
|--------|------------------------------|----------|------|--|
| Symbol | | | | |
| | MIN | TYP | MAX | |
| Α | - | - | 1.20 | |
| A1 | 0.05 | - | 0.15 | |
| A2 | 0.90 | 1.00 | 1.05 | |
| b | 0.20 | - | 0.28 | |
| С | 0.10 | - | 0.19 | |
| D | 4.86 | 4.96 | 5.06 | |
| Е | 6.20 | 6.40 | 6.60 | |
| E1 | 4.30 | 4.40 | 4.50 | |
| е | | 0.65 BSC | | |
| L | 0.45 0.60 0.75 | | | |
| L1 | 1.00 REF | | | |
| L2 | 0.25 BSC | | | |
| R | 0.09 | | | |
| θ | 0° | - | 8° | |

1.8V, 34µA, RRIO, Zero Drift Op-amps Package Outline Dimensions

SOP-14 (SOIC-14)



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HA1630S01LPEL-E SCY33178DR2G NJU77806F3-TE1 NCV5652MUTWG NCV20034DR2G LM2902EDR2G NTE778S NTE871

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LM4565FVT-GE2 EL5420CRZ-T7A TSV791IYLT TSV772IQ2T AS324AMTR-E1 TLV2772QPWR NJM4556AM-TE1 NJM2068M-TE1

AS324MTR-E1 AS358MMTR-G1 MCP6232T-EMNY MCP662-E/MF TLC081AIP TLC082AIP TLE2074ACDW TLV07IDR

TLV2170IDGKT