## International IgR Rectifier

$\mathrm{Reference}^{D_{\text {esign }}}$IRDCiP2005C-2

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IRDCiP2005C-2: 500kHz, 30A, Dual Output, $180^{\circ}$ Out of Phase Synchronous Buck Converter Featuring iP2005C and IR3623M

## Overview

This reference design is capable of delivering a continuous current of 30A per channel without heatsink at an ambient temperature of $45^{\circ} \mathrm{C}$ and airflow of 200LFM. Fig. 4 - Fig. 25 provide performance graphs, thermal images, and waveforms. Fig. 1 - Fig. 3 are provided to engineers as design references for implementing an IR3623+iP2005C solution.
The components installed on this demoboard were selected based on operation at an input voltage of $12 \mathrm{~V}(+/-10 \%)$, a switching frequency of $500 \mathrm{kHz}(+/-$ $15 \%$ ), and an output voltage of 1.5 V at channel -1 and 2.5 V at channel- 2 . Major
 changes from these set points may require optimizing the control loop and/or adjusting the values of input/output filters in order to meet the user's specific application requirements. Refer to iP2005C and IR3623 datasheets for more information.

## IRDCiP2005C-2 Recommended Operating Conditions

(refer to the iP2005C datasheet for maximum operating conditions)

Input voltage:
Output voltage:
$8.5 \mathrm{~V}-14.5 \mathrm{~V}$

Switching Freq:
Output current:
$0.8-5 \mathrm{~V}$
500 kHz

This reference design is capable of delivering a continuous current of 30A per channel without heatsink at an ambient temperature of $45^{\circ} \mathrm{C}$ and airflow of 200LFM.

## Demoboard Quick Start Guide

## Initial Settings:

VOUT1 is set to 1.5 V , but can be adjusted from 0.8 V to 5 V by changing the values of R 11 and R 15 according to the following formula:

$$
\text { R11 }=\text { R15 }=(10 \mathrm{k} * 0.8) /(\text { VOUT1 - 0.8 })
$$

VOUT2 is set to 2.5 V , but can be adjusted from 0.8 V to 5 V by changing the values of R 12 and R 16 according to the following formula:
R12 = R16 = (10k * 0.8) / (VOUT2 - 0.8)

The switching frequency is set to 500 kHz , but can be adjusted by changing the value of R26. See Fig. 4 for the relationship between R26 and the switching frequency.

Power Up Procedure:

1. Apply input voltage across VIN and PGND.
2. If R45 is not installed, apply bias voltage across VDD and PGND.
3. Apply load across VOUT pads and PGND pads.
4. Toggle the SEQ (SW1) and EN (SW2) switches to the ON position.
5. Adjust load to desired level. See recommendations above.

## Demoboard Schematic



Fig. 1 Schematic

## International IgRRectifier

| Quantity | Designator | Type 1 | Type 2 | Value 1 | Value 2 | Tolerance | Package | Manufac 1 | Manufac 1No |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 20 | C1, C2, C3, C4, C5, C6, C7, C8, C9, C10, C11, C12, C17, C18, C21, C22, C25, C26, C29, C30 | capacitor | X7R | 10.0uF | 16 V | 10\% | 1206 | TDK | C3216X7R1C106KT |
| 3 | C13, C33, C34 | capacitor | X7R | 0.100 uF | 50 V | 10\% | 0603 | TDK | C1608X7R1H104K |
| 1 | C14 | capacitor | electrolytic | 680uF | 16 V | 20\% | SMD | Panasonic | EEV-FK1C681GP |
| 8 | C15, C16, C19, C20, C23, C24, C27, C28 | capacitor | X5R | 100uF | 6.3 V | 20\% | 1210 | TDK | C3225X5R0J107M |
| 11 | $\begin{gathered} \text { C31, C43, C44, C45, C46, C47, C48, C53, } \\ \text { C54, C55, C56 } \end{gathered}$ | capacitor | X7R | 1.00uF | 16 V | 10\% | 0603 | TDK | C1608X7R1C105KT |
| 5 | C32, C41, C42, C51, C52 | capacitor | NPO | 100pF | 50 V | 5\% | 0603 | Phycomp | 0603CG101J9B20 |
| 2 | C35, C36 | capacitor | NPO | 47.0pF | 50 V | 5\% | 0603 | KOA | NPO0603HTTD470J |
| 2 | C37, C38 | capacitor | X7R | 2700 pF | 50 V | 10\% | 0603 | KOA | X7R0603HTTD272K |
| 2 | C39, C40 | capacitor | NPO | 390pF | 50 V | 5\% | 0603 | KOA | NPO0603HTTD391J |
| 4 | C59, C60, C63, C64 | capacitor | tantalum polymer | 220uF | 2.5 V | 20\% | 7343 | Sanyo | 2R5TPC220M |
| 2 | L1, L2 | inductor | ferrite | 0.22 uH | 47A | 10\% | SMT | Vitec | 59PR9873N |
| 10 | $\begin{gathered} \hline \text { R1, R2, R9, R10, R13, R14, R23, R24, R33, } \\ \text { R34 } \end{gathered}$ | resistor | thick film | 10.0K | 1/10W | 1\% | 0603 | KOA | RK73H1J1002F |
| 2 | R11, R15 | resistor | thick film | 11.5K | 1/10W | 1\% | 0603 | KOA | RK73H1JLTD1152F |
| 2 | R12, R16 | resistor | thick film | 4.75 K | 1/10W | 1\% | 0603 | KOA | RK73H1JLTD4751F |
| 2 | R17, R18 | resistor | thick film | 3.65 K | 1/10W | 1\% | 0603 | KOA | RK73H1JLTD3651F |
| 2 | R19, R20 | resistor | thick film | 0 | 1/8W | <50m | 0805 | ROHM | MCR10EZHJ000 |
| 1 | R25 | resistor | thick film | 30.1 K | 1/10W | 1\% | 0603 | KOA | RK73H1J3012F |
| 1 | R26 | resistor | thick film | 78.7 K | 1/10W | 1\% | 0603 | KOA | RK73H1JLTD7872F |
| 12 | R3, R4, R27, R28, R31, R32, R36, R37, R38, R47, R48, R50 | resistor | thick film | 0 | 1/10W | 1\% | 0603 | KOA | RK73Z1JLTD |
| 1 | R35 | resistor | thick film | 0 | 1/8W | <50m | 1206 | Panasonic | ERJ-8GEY0R00 |
| 2 | R5, R6 | resistor | thick film | 11.0K | 1/10W | 1\% | 0603 | KOA | RK73H1JLTD1102F |
| 2 | R7, R8 | resistor | thick film | 200 | 1/10W | 1\% | 0603 | KOA | RK73H1J2000F |
| 2 | SW1, SW2 | switch | slide | SPDT | 30VDC | 0.2 A | pcb mount | E-Switch | EG1218 |
| 18 | TP1, TP2, TP3, TP4, TP5, TP6, TP7, TP8, TP27, TP28, TP29, TP30, TP31, TP32, TP33, TP34, TP35, TP36 | hardware | test point | 90 mils | $112 \times 150$ mils | - | 5016 | Keystone | 5016 |
| 20 | TP9, TP10, TP11, TP12, TP13, TP14, TP15, TP16, TP17, TP18, TP19, TP20, TP21, TP22, TP23, TP24, TP25, TP26, TP37, TP38 | hardware | test point | 60 mils | $40 \times 105$ mils | - | 5015 | Keystone | 5015 |
| 2 | U1, U2 | iP2005 | LGA unit | reve | - | - | $7.65 \mathrm{~mm} \times 7.65 \mathrm{~mm}$ | IRF | reve |
| 1 | U3 | IC analog | PWM controller | -0.5-16V | -0.5-16V | $-40-120^{\circ} \mathrm{C}$ | MLPQ-32L | IRF | IR3623M |

## Demoboard Component Placement



Fig. 2 Top Layer (Face View)


Fig. 3 Bottom Layer (Through View)

Description of Test Points and Connectors

1. Jumpers

| Jumper | Pin Name | Description |
| :---: | :--- | :--- |
| SW1 | EN | Board Enable ( switch Up=Off, Down=On ) - Vin pin on top |
| SW2 | SEQ | Sequence ( switch Up=Off, Down=On ) - Vin pin on top |

2. Test Points/Connectors

| Test Point | Pin Name | Description |
| :---: | :---: | :---: |
| T1 / T2 | VIN / PGND | Vin supply voltage |
| TP2 / TP28 | VIN / PGND | Vin supply voltage sense |
| T3 / T5 / T7 | VOUT1 / PGND / PGND | Channel 1 Output, connect to DC load |
| TP35 / TP33 | VOUT1 / PGND | Channel 1 Output sense |
| TP21 / TP37 | VSW1 / PGND | Channel 1 switch node / PGND test points |
| TP9 | EN1 | Channel 1 Enable test point |
| TP11 | PWM1 | Channel 1 PWM test point |
| TP19 | CC1 | Channel 1 error amplifier output |
| TP25 | FB1 | Channel 1 error amplifier non-inverting input |
| T4 / T6 / T9 | VOUT2 / PGND / PGND | Channel 2 Output, connect to DC load |
| TP36 / TP34 | VOUT2 / PGND | Channel 2 Output sense |
| TP22 / TP38 | VSW2 / PGND | Channel 2 switch node / PGND test points |
| TP10 | EN2 | Channel 2 Enable test point |
| TP12 | PWM2 | Channel 2 PWM test point |
| TP20 | CC2 | Channel 2 error amplifier output |
| TP26 | FB2 | Channel 2 error amplifier non-inverting input |
| TP7 / TP8 | VDD / PGND | iP2005C internal bias voltage test points |
| TP23 | SYNC | External frequency synchronization input |
| TP17 | TRACK1 | Channel 1 tracking input, pull-up to Vout3 if not used |
| TP18 | TRACK2 | Track2 test point |
| TP15 | PGOOD1 | Channel 1 Power good test point |
| TP16 | PGOOD2 | Channel 2 Power good test point |
| TP13 | SS1 | Channel 1 Soft start test point |
| TP14 | SS2 | Channel 2 Soft start test point |
| TP24 | FAULT | Fault monitor test point |

3. Test points for Efficiency Measurement

| Test Point | Pin Name | Description |
| :---: | :--- | :--- |
| TP1 / TP4 | +VINS1 / -VOUTS1 | Channel 1 Vin sense for efficiency measurement |
| TP3 / TP4 | +VOUTS1 / -VOUTS1 | Channel 1 Output sense for efficiency measurement |
| TP27 / TP6 | +VINS2 / -VOUTS2 | Channel 2 Vin sense for efficiency measurement |
| TP5 / TP6 | +VOUTS2 / -VOUTS2 | Channel 2 Output sense for efficiency measurement |

## Test Results



Fig. 4 Relationship Between Switching Frequency and R26


Fig. 5 Channel-1 Power Up Sequence (C3: EN, C1: SS1, C4: VOUT1)


Fig. 6 Channel-1 Power Down Sequence (C3: EN, C1: SS1, C4: VOUT1)


Fig. 7 Channel-2 Power Up Sequence (C3: EN, C1: SS2, C4: VOUT2)


Fig. 8 Channel-2 Power Down Sequence (C3: EN, C1: SS2, C4: VOUT2)


Fig. 9 Hiccup Mode Over Current Protection (C1: SS1, C4: $\mathrm{I}_{\text {out1 }}, \mathrm{C} 3:$ VOUT1)


Fig. 10 Hiccup Mode Over Current Protection (C1: SS1, C4: $\mathrm{I}_{\mathrm{out} 1}$, C3: VOUT1)


Fig. 11 Deadtime and Ringing at Switch Node


Fig. 12 Channel-1 Output Voltage DC Ripple


Fig. 13 Channel-1 Output Voltage DC Ripple


Fig. 14 Channel-2 Output Voltage DC Ripple


Fig. 15 Channel-2 Output Voltage DC Ripple


Fig. 16 Load Transient Response (C1: VOUT1 - AC, C2: $\mathrm{I}_{\text {out1 }}$ divided by 2)


Fig. 17 Load Transient Response (C1: VOUT2 - AC, C2: $\mathrm{I}_{\mathrm{out} 2}$ divided by 2)


Phase (deg)
153.00
126.00
99.00
72.00
45.00
18.00
-9.00
$-36.00$
-63.00

Fig. 18 Bode Plot (VIN = 12V, VOUT1 = 1.5V, $\mathrm{I}_{\text {out }}=20 \mathrm{~A}$ )


Fig. 19 Bode Plot (VIN = 12V, VOUT2 = 2.5V, $\mathrm{I}_{\text {out } 2}=20 \mathrm{~A}$ )


Fig. 20 Channel-1 Power Loss


Fig. 21 Channel-1 Efficiency


Fig. 22 Channel-2 Power Loss


Fig. 23 Channel-2 Efficiency


Fig. 24 Thermal Image: $\mathrm{I}_{\text {out }}=30 \mathrm{~A}$ per channel, VIN $=12 \mathrm{~V}$, with Internal VDD, VOUT1 $=1.5 \mathrm{~V}, \mathrm{VOUT}=2.5 \mathrm{~V}$, $\mathrm{TA}=45^{\circ} \mathrm{C}, \mathrm{f}_{\mathrm{sw}}=500 \mathrm{kHz}, 200 \mathrm{LFM}$, No Heatsink


Fig. 25 Thermal Image: $\mathrm{I}_{\text {out }}=30 \mathrm{~A}$ per channel, $\mathrm{VIN}=12 \mathrm{~V}, \mathrm{VDD}=6 \mathrm{~V}, \mathrm{VOUT} 1=1.5 \mathrm{~V}, \mathrm{VOUT} 2=2.5 \mathrm{~V}, \mathrm{TA}=$ $45^{\circ} \mathrm{C}, \mathrm{f}_{\mathrm{sw}}=500 \mathrm{kHz}, 200 \mathrm{LFM}$, No Heatsink

Table 1 Maximum Temperature for iP2005C Dual Output Configuration

| Bias Voltage | U 1 | U 2 |
| :---: | :---: | :---: |
| Internal VDD $=5.2 \mathrm{~V}$ | $103^{\circ} \mathrm{C}$ | $108^{\circ} \mathrm{C}$ |
| External VDD $=6 \mathrm{~V}$ | $101^{\circ} \mathrm{C}$ | $106^{\circ} \mathrm{C}$ |

## IRDCiP2005C-2

Refer to the following application notes for detailed guidelines and suggestions when implementing iPOWIR Technology products:

## AN-1043: Stabilize the Buck Converter with Transconductance Amplifier

This paper explains how to design the voltage compensation network for Buck Converters with Transconductance Amplifier. The design methods and equations for Type II and Type III compensation are given.

AN-1028: Recommended Design, Integration and Rework Guidelines for International Rectifier's iPowIR Technology BGA and LGA and Packages
This paper discusses optimization of the layout design for mounting iPowIR BGA and LGA packages on printed circuit boards, accounting for thermal and electrical performance and assembly considerations. Topics discussed include PCB layout placement, and via interconnect suggestions, as well as soldering, pick and place, reflow, inspection, cleaning and reworking recommendations.

## AN-1030: Applying iPOWIR Products in Your Thermal Environment

This paper explains how to use the Power Loss and SOA curves in the data sheet to validate if the operating conditions and thermal environment are within the Safe Operating Area of the iPOWIR product.

## AN-1047: Graphical solution for two branch heatsinking Safe Operating Area

Detailed explanation of the dual axis SOA graph and how it is derived.
Use of this design for any application should be fully verified by the customer. International Rectifier cannot guarantee suitability for your applications, and is not liable for any result of usage for such applications including, without limitation, personal or property damage or violation of third party intellectual property rights.

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