

PIC16F882/883/884/886/887 Data Sheet

28/40/44-Pin, Enhanced Flash-Based 8-Bit CMOS Microcontrollers with nanoWatt Technology

Preliminary

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28/40/44-Pin Flash-Based, 8-Bit CMOS Microcontrollers with nanoWatt Technology

High-Performance RISC CPU:

- Only 35 instructions to learn:
 - All single-cycle instructions except branches
- Operating speed:
 - DC 20 MHz oscillator/clock input
 - DC 200 ns instruction cycle
- · Interrupt capability
- 8-level deep hardware stack
- Direct, Indirect and Relative Addressing modes

Special Microcontroller Features:

- Precision Internal Oscillator:
 - Factory calibrated to ±1%
 - Software selectable frequency range of 8 MHz to 31 kHz
 - Software tunable
 - Two-Speed Start-up mode
 - Crystal fail detect for critical applications
 - Clock mode switching during operation for power savings
- Power-Saving Sleep mode
- Wide operating voltage range (2.0V-5.5V)
- Industrial and Extended Temperature range
- Power-on Reset (POR)
- Power-up Timer (PWRT) and Oscillator Start-up Timer (OST)
- Brown-out Reset (BOR) with software control option
- Enhanced low-current Watchdog Timer (WDT) with on-chip oscillator (software selectable nominal 268 seconds with full prescaler) with software enable
- Multiplexed Master Clear with pull-up/input pin
- Programmable code protection
- High Endurance Flash/EEPROM cell:
- 100,000 write Flash endurance
- 1,000,000 write EEPROM endurance
- Flash/Data EEPROM retention: > 40 years
- Program memory Read/Write during run time
- In-Circuit Debugger (on board)

Low-Power Features:

- Standby Current:
- 50 nA @ 2.0V, typical
- Operating Current:
- 11 μA @ 32 kHz, 2.0V, typical
- 220 μA @ 4 MHz, 2.0V, typical
- Watchdog Timer Current:
 - 1 μA @ 2.0V, typical

Peripheral Features:

- 24/35 I/O pins with individual direction control:
 - High current source/sink for direct LED drive
 - Interrupt-on-Change pin
 - Individually programmable weak pull-ups
 - Ultra Low-Power Wake-up (ULPWU)
- Analog Comparator module with:
 - Two analog comparators
 - Programmable on-chip voltage reference (CVREF) module (% of VDD)
 - Fixed voltage reference (0.6V)
 - Comparator inputs and outputs externally accessible
 - SR Latch mode
 - External Timer1 Gate (count enable)
- A/D Converter:
 - 10-bit resolution and 11/14 channels
- Timer0: 8-bit timer/counter with 8-bit programmable prescaler
- Enhanced Timer1:
- 16-bit timer/counter with prescaler
- External Gate Input mode
- Dedicated low-power 32 kHz oscillator
- Timer2: 8-bit timer/counter with 8-bit period register, prescaler and postscaler
- Enhanced Capture, Compare, PWM+ module:
 - 16-bit Capture, max. resolution 12.5 ns
 - Compare, max. resolution 200 ns
 - 10-bit PWM with 1, 2 or 4 output channels, programmable "dead time", max. frequency 20 kHz
 - PWM output steering control
- Capture, Compare, PWM module:
 - 16-bit Capture, max. resolution 12.5 ns
 - 16-bit Compare, max. resolution 200 ns
 - 10-bit PWM, max. frequency 20 kHz
- Enhanced USART module:
 - Supports RS-485, RS-232, and LIN 2.0Auto-Baud Detect
 - Auto-Wake-Up on Start bit
- In-Circuit Serial Programming[™] (ICSP[™]) via two pins
- Master Synchronous Serial Port (MSSP) module supporting 3-wire SPI (all 4 modes) and I²C[™] Master and Slave Modes with I²C address mask

Davias	Program Memory	Data N	lemory	1/0	10-bit A/D	ECCP/	EUGADT	Meed	Comporatora	Timers
Device	Flash (words)	SRAM (bytes)	EEPROM (bytes)	S) (ch) CCP CCP		EUSARI			8/16-bit	
PIC16F882	2048	128	128	28	11	1/1	1	1	2	2/1
PIC16F883	4096	256	256	24	11	1/1	1	1	2	2/1
PIC16F884	4096	256	256	35	14	1/1	1	1	2	2/1
PIC16F886	8192	368	256	24	11	1/1	1	1	2	2/1
PIC16F887	8192	368	256	35	14	1/1	1	1	2	2/1

Pin Diagrams – PIC16F882/883/886, 28-Pin PDIP, SOIC, SSOP

28-pin PDIP, SOIC, SSOP ° 1 RE3/MCLR/VPP -28 **RB7/ICSPDAT** RA0/AN0/ULPWU/C12IN0- -2 27 **RB6/ICSPCLK** RA1/AN1/C12IN1-3 26 RB5/AN13/T1G 4 25 RA2/AN2/VREF-/CVREF/C2IN+ -RB4/AN11/P1D PIC16F882/883/886 5 24 RB3/AN9/PGM/C12IN2-RA3/AN3/VREF+/C1IN+ < RA4/T0CKI/C1OUT --RB2/AN8/P1B 6 23 RA5/AN4/SS/C2OUT -Г 7 22 RB1/AN10/P1C/C12IN3-RB0/AN12/INT -Vss 8 21 · RA7/OSC1/CLKIN < 9 20 Vdd -RA6/OSC2/CLKOUT -10 19 Vss RC0/T1OSO/T1CKI -11 18 RC7/RX/DT RC1/T1OSI/CCP2 --12 17 RC6/TX/CK RC5/SDO RC2/P1A/CCP1 --13 16 RC4/SDI/SDA RC3/SCK/SCL -14 15

TABLE 1:PIC16F882/883/886 28-PIN SUMMARY (PDIP, SOIC, SSOP)

I/O	Pin	Analog	Comparators	Timers	ECCP	EUSART	MSSP	Interrupt	Pull-up	Basic
RA0	2	AN0/ULPWU	C12IN0-	_	—	—	—	—	—	_
RA1	3	AN1	C12IN1-	_	_	_	_	_	_	_
RA2	4	AN2	C2IN+	—	—	_	_	_	_	VREF-/CVREF
RA3	5	AN3	C1IN+	—	—	-	_	_	_	Vref+
RA4	6	—	C1OUT	TOCKI	—					_
RA5	7	AN4	C2OUT		_		s			
RA6	10	_	_		_					OSC2/CLKOUT
RA7	9	_	_		_					OSC1/CLKIN
RB0	21	AN12	_	_	—	_	_	IOC/INT	Y	_
RB1	22	AN10	C12IN3-		P1C			IOC	Y	
RB2	23	AN8	_		P1B			IOC	Y	
RB3	24	AN9	C12IN2-		_			IOC	Y	PGM
RB4	25	AN11	_		P1D			IOC	Y	
RB5	26	AN13	—	T1G	_			IOC	Y	
RB6	27	—	—	—	—	_	—	IOC	Y	ICSPCLK
RB7	28	_	—		_			IOC	Y	ICSPDAT
RC0	11	—	—	T1OSO/T1CKI	—	_	_	_	_	_
RC1	12	_	—	T1OSI	CCP2	_	_	_	_	—
RC2	13	_	_		CCP1/P1A					
RC3	14	_	—		_		SCK/SCL			
RC4	15	-	_		-		SDI/SDA			
RC5	16	_	—		_		SDO			
RC6	17	_	_		_	TX/CK				
RC7	18	_	—		—	RX/DT				
RE3	1	—	_	_	—	—	—	_	Y(1)	MCLR/VPP
—	20	—	—		_	_	_	_	_	VDD
_	8	—	—		_	_	_	_	—	Vss
—	19	_	—		_	_	_	_	_	Vss

Note 1: Pull-up activated only with external MCLR configuration.

Pin Diagrams – PIC16F882/883/886, 28-Pin QFN



I/O	Pin	Analog	Comparators	Timers	ECCP	EUSART	MSSP	Interrupt	Pull-up	Basic
RA0	27	AN0/ULPWU	C12IN0-		_	_	_	_	_	—
RA1	28	AN1	C12IN1-		_	_	_	_	_	_
RA2	1	AN2	C2IN+		_	—	—	_	—	VREF-/CVREF
RA3	2	AN3	C1IN+		_	_	_			VREF+
RA4	3	_	C1OUT	T0CKI	_	_	_			—
RA5	4	AN4	C2OUT		—	—	SS	—	—	—
RA6	7	—	—		—	_	_	_	_	OSC2/CLKOUT
RA7	6	—	_		—	_	—	_	_	OSC1/CLKIN
RB0	18	AN12	_	—	_	—		IOC/INT	Y	_
RB1	19	AN10	C12IN3-		P1C		—	IOC	Y	_
RB2	20	AN8		—	P1B	—	—	IOC	Y	_
RB3	21	AN9	C12IN2-	—	—	—	—	IOC	Y	PGM
RB4	22	AN11	—	—	P1D	—	_	IOC	Y	—
RB5	23	AN13	—	T1G	—	—	—	IOC	Y	—
RB6	24	—	—	—	—	—	_	IOC	Y	ICSPCLK
RB7	25	—	—		—	—	—	IOC	Y	ICSPDAT
RC0	8	—	—	T1OSO/T1CKI	—	—	_	_	—	—
RC1	9	—	—	T1OSI	CCP2	—	—	_	—	—
RC2	10	—	—		CCP1/P1A	—	_	_	—	—
RC3	11	—	_		—	_	SCK/SCL	_	_	—
RC4	12	—	_	_	—	—	SDI/SDA	—	—	_
RC5	13	—	—		—	—	SDO	_	—	—
RC6	14	—	_		—	TX/CK	_	_	_	—
RC7	15	—	_	_	—	RX/DT	_	_	—	_
RE3	26	_	_	_	—	_		_	Y(1)	MCLR/VPP
	17	_	—		—	—	_	_	_	Vdd
_	5	—	_		—	—	—	—	_	Vss
—	16	—	—	—	—	—	—	—	—	Vss

TABLE 2:	PIC16F882/883/886 28-PIN SUMMARY (QFN)
		wiit)

Note 1: Pull-up activated only with external $\overline{\text{MCLR}}$ configuration.

Pin Diagrams - PIC16F884/887, 40-Pin PDIP

40-pin PDIP



IABL	E 3: PIC16F884/887 40-PIN SUMMARY (PDIP)									
I/O	Pin	Analog	Comparators	Timers	ECCP	EUSART	MSSP	Interrupt	Pull-up	Basic
RA0	2	AN0/ULPWU	C12IN0-	_	_		_	_		_
RA1	3	AN1	C12IN1-	_	_	_	_	_		—
RA2	4	AN2	C2IN+	_	_	_	_	_		VREF-/CVREF
RA3	5	AN3	C1IN+	_	_	_	_	_	_	Vref+
RA4	6	_	C1OUT	T0CKI	_	_	_			_
RA5	7	AN4	C2OUT	_	_	_	SS	_	_	—
RA6	14	_	—	_	_	_	_	_		OSC2/CLKOUT
RA7	13	_	—	_	_	_	_	_	_	OSC1/CLKIN
RB0	33	AN12	—	—	—	-	—	IOC/INT	Y	—
RB1	34	AN10	C12IN3-	—	—	-	—	IOC	Y	—
RB2	35	AN8	—	_	_	_	_	IOC	Y	—
RB3	36	AN9	C12IN2-		—		_	IOC	Y	PGM
RB4	37	AN11	—	—	—	-	—	IOC	Y	—
RB5	38	AN13	—	T1G	—	-	—	IOC	Y	—
RB6	39	—	—		—		—	IOC	Y	ICSPCLK
RB7	40	—	—		_		_	IOC	Y	ICSPDAT
RC0	15	—	_	T1OSO/T1CKI	_		_	—		_
RC1	16	—	_	T1OSI	CCP2		_	_		_
RC2	17	—	_		CCP1/P1A		_	—		_
RC3	18	—	—		_		SCK/SCL	—		—
RC4	23	—	_		_		SDI/SDA	_		_
RC5	24	—		_	—	_	SDO	_		_
RC6	25	—	_	_	—	TX/CK	—	—	_	
RC7	26	—		_	—	RX/DT	—	_		_
RD0	19	—		—	—	_	—	—	—	
RD1	20	—	_	_	—	_	_	_	—	_
RD2	21	—	—	_	_	_	_	—		—
RD3	22	—		_	—	_	—	—	—	
RD4	27	_			_	_	_	—	—	
RD5	28	—			P1B	_	_	_	—	
RD6	29	_		_	P1C	_			—	_
RD7	30	—			P1D	_	_	_	—	
RE0	8	AN5						—		
RE1	9	AN6				_		_	—	
RE2	10	AN7						—		
RE3	1	—	_	_	_	-	_	_	Y(1)	MCLR/Vpp
	11	—	—	_	—	_	_	—	—	Vdd
	32								—	Vdd
_	12	_	_	_	_	_	_	_	_	Vss
_	31	—			_	_	_	_	_	Vss

Note 1: Pull-up activated only with external MCLR configuration.

Pin Diagrams – PIC16F884/887, 44-Pin QFN



IABL	E 4:	4: PIC16F884/887 44-PIN SUMMARY (QFN)								
I/O	Pin	Analog	Comparators	Timers	ECCP	EUSART	MSSP	Interrupt	Pull-up	Basic
RA0	19	AN0/ULPWU	C12IN0-	—	—	_	_	_		—
RA1	20	AN1	C12IN1-		_	_	_			—
RA2	21	AN2	C2IN+	—	_		_			VREF-/CVREF
RA3	22	AN3	C1IN+	_			_			Vref+
RA4	23	—	C1OUT	TOCKI	_	—	_	_		_
RA5	24	AN4	C2OUT		_	_	SS	_		_
RA6	33	_	_	—	_		_			OSC2/CLKOUT
RA7	32	_		_			_			OSC1/CLKIN
RB0	9	AN12	—	—	_	_	_	IOC/INT	Y	—
RB1	10	AN10	C12IN3-	—	_	_	_	IOC	Y	—
RB2	11	AN8	—	—	_	_	_	IOC	Y	—
RB3	12	AN9	C12IN2-	—	—	—	-	IOC	Y	PGM
RB4	14	AN11	—	—	_	_	_	IOC	Y	—
RB5	15	AN13	—	T1G	_	_	_	IOC	Y	—
RB6	16	_	—		_		_	IOC	Y	ICSPCLK
RB7	17	_	—	—	_	_	_	IOC	Y	ICSPDAT
RC0	34	_	_	T1OSO/T1CKI	_		_			—
RC1	35	_		T1OSI	CCP2		_			—
RC2	36	_	—	—	CCP1/P1A	_	_			—
RC3	37	_		_			SCK/SCL			—
RC4	42	_	_	—	_		SDI/SDA			—
RC5	43	_	—	—	_	_	SDO	_	_	—
RC6	44	_	—	—	_	TX/CK	_			—
RC7	1	_	—	—	_	RX/DT	_	_	_	—
RD0	38	—	—	—	—	—	—	—		—
RD1	39	—	—	—	—	—	-	_		—
RD2	40	—	—	—	—	—		_		—
RD3	41	—	—	—	—	—		—		—
RD4	2	—	—	—	—	—	_	—	—	—
RD5	3	—		_	P1B	—	_	_		
RD6	4	—	_	_	P1C	_		_		—
RD7	5	—		_	P1D	—	_	_		
RE0	25	AN5	—	—	—	—	_	—		—
RE1	26	AN6		_	—	—	_	—		_
RE2	27	AN7	—	—	—	—	_	—		—
RE3	18	—	—	—	—	—	_		Y(1)	MCLR/VPP
—	7	—	—	—	—	—	—	—	—	Vdd
—	8	—		_	—	—	_	_		Vdd
—	28	—	—	_	—	—		_		Vdd
	6	_	_		_		_	_		Vss
	30	_	_				_		_	Vss
	31	_	_		_		_	_		Vss
	13	_	_		_	—	_	_		NC (no connect)
	29	—	_		_	_	_		_	NC (no connect)

TABLE 4:	PIC16F884/887	44-PIN	SUMMARY	(QFN)

Note 1: Pull-up activated only with external MCLR configuration.

Pin Diagrams - PIC16F884/887, 44-Pin TQFP



IABL	E 5:	PIC16	F884/887 44	I-PIN SUMMA	KY (IQFI	י)				
I/O	Pin	Analog	Comparators	Timers	ECCP	EUSART	MSSP	Interrupt	Pull-up	Basic
RA0	19	AN0/ULPWU	C12IN0-	—	—	_	_	—	—	—
RA1	20	AN1	C12IN1-	—	_	_		_		—
RA2	21	AN2	C2IN+	—	—	_	—	—	—	VREF-/CVREF
RA3	22	AN3	C1IN+	—	—	—	_	—	—	VREF+
RA4	23	_	C1OUT	T0CKI	_		_	_		_
RA5	24	AN4	C2OUT	—	—	—	SS	—	—	—
RA6	31	—	—	—	—	_	—	—	—	OSC2/CLKOUT
RA7	31	—	—	—	—	—		—	—	OSC1/CLKIN
RB0	8	AN12	—	_	_			IOC/INT	Y	—
RB1	9	AN10	C12IN3-	—	—	_		IOC	Y	—
RB2	10	AN8	—	—	_	_	_	IOC	Y	—
RB3	11	AN9	C12IN2-	—	—	_	-	IOC	Y	PGM
RB4	14	AN11	—	—	—	_	—	IOC	Y	—
RB5	15	AN13	—	T1G	_	_	_	IOC	Y	—
RB6	16	—	—	—	—		_	IOC	Y	ICSPCLK
RB7	17	_	—	—	_	_	_	IOC	Y	ICSPDAT
RC0	32	—	_	T1OSO/T1CKI	_		_			
RC1	35		_	T1OSI	CCP2	_	_	_		_
RC2	36	—	—	—	CCP1/P1A		_	_	_	—
RC3	37	_		_	_		SCK/SCL	_		
RC4	42	—	_	—	_		SDI/SDA			
RC5	43	_	—	—	_	_	SDO	_	—	—
RC6	44	_	—	—	_	TX/CK	_	_		—
RC7	1	_		_	_	RX/DT	_	_		
RD0	38	—	—	—	—		_	—	—	—
RD1	39	—	—	—	—	—	_	—	—	—
RD2	40	—	—	—	—	_	—	—	—	—
RD3	41	—	—	—	—	—		—	—	—
RD4	2	—	—	—	_			—	—	—
RD5	3	—	—	—	P1B	_		—	—	—
RD6	4	—	_	_	P1C	_		_	_	—
RD7	5	—		_	P1D	_	_	_	—	—
RE0	25	AN5	—	—	—	_	_	—	—	—
RE1	26	AN6	_	_	—	_		_	_	—
RE2	27	AN7	—	—	—		_	—	—	—
RE3	18	—	—	—	—	_		—	Y(1)	MCLR/VPP
—	7	—	_	_	—	_		_	_	Vdd
—	28	—	—	—	—	_		—	—	Vdd
	6	_	_		_	_		_	_	Vss
	13	_	_	_	_			_		NC (no connect)
	29	_	_		_	_		_		Vss
_	34	_	_		_					NC (no connect)
	33	_	_		—	_				NC (no connect)
_	12	_	_		_	_	_	_	_	NC (no connect)

TABLE 5:	PIC16F884/887 44-PIN SUMMARY ((TQFP)

Note 1: Pull-up activated only with external MCLR configuration.

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1.0 DEVICE OVERVIEW

The PIC16F882/883/884/886/887 is covered by this data sheet. The PIC16F882/883/886 is available in 28pin PDIP, SOIC, SSOP and QFN packages. The PIC16F884/887 is available in a 40-pin PDIP and 44pin QFN and TQFP packages. Figure 1-1 shows the block diagram of PIC16F882/883/886 and Figure 1-2 shows a block diagram of the PIC16F884/887 device. Table 1-1 and Table 1-2 show the corresponding pinout descriptions.







TABLE 1-1: PIC16F882/883/886 PINOUT DESCRIPTION

Name	Function	Input Type	Output Type	Description
RA0/AN0/ULPWU/C12IN0-	RA0	TTL	CMOS	General purpose I/O.
	AN0	AN		A/D Channel 0 input.
	ULPWU	AN		Ultra Low-Power Wake-up input.
	C12IN0-	AN		Comparator C1 or C2 negative input.
RA1/AN1/C12IN1-	RA1	TTL	CMOS	General purpose I/O. Individually enabled pull-up.
	AN1	AN		A/D Channel 1 input.
	C12IN1-	AN		Comparator C1 or C2 negative input.
RA2/AN2/VREF-/CVREF/C2IN+	RA2	TTL	CMOS	General purpose I/O.
	AN2	AN	_	A/D Channel 2.
	VREF-	AN		A/D Negative Voltage Reference input.
	CVREF	_	AN	Comparator Voltage Reference output.
	C2IN+	AN		Comparator C2 positive input.
RA3/AN3/VREF+/C1IN+	RA3	TTL		General purpose I/O.
	AN3	AN		A/D Channel 3.
	VREF+	AN	_	Programming voltage.
	C1IN+	AN	_	Comparator C1 positive input.
RA4/T0CKI/C1OUT	RA4	TTL	CMOS	General purpose I/O. Individually enabled pull-up.
	T0CKI	ST	_	Timer0 clock input.
	C10UT	—	CMOS	Comparator C1 output.
RA5/AN4/SS/C2OUT	RA5	TTL	CMOS	General purpose I/O.
	AN4	AN	_	A/D Channel 4.
	SS	ST	_	Slave Select input.
	C2OUT	—	CMOS	Comparator C2 output.
RA6/OSC2/CLKOUT	RA6	TTL	CMOS	General purpose I/O.
	OSC2	—	XTAL	Master Clear with internal pull-up.
	CLKOUT	_	CMOS	Fosc/4 output.
RA7/OSC1/CLKIN	RA7	TTL	CMOS	General purpose I/O.
	OSC1	XTAL		Crystal/Resonator.
	CLKIN	ST		External clock input/RC oscillator connection.
RB0/AN12/INT	RB0	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	AN12	AN		A/D Channel 12.
	INT	ST		External interrupt.
RB1/AN10/P1C/C12IN3-	RB1	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	AN10	AN		A/D Channel 10.
	P1C	_	CMOS	PWM output.
	C12IN3-	AN		Comparator C1 or C2 negative input.
RB2/AN8/P1B	RB2	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	AN8	AN	—	A/D Channel 8.
	P1B	—	CMOS	PWM output.
Legend: AN = Analog input TTL = TTL compatil HV = High Voltage	or output ble input	CMOS ST XTAL	= CMO = Schm = Crysta	S compatible input or output OD = Open Drain itt Trigger input with CMOS levels al

TABLE 1-1: PIC16F882/883/886 PINOUT DESCRIPTION (CONTINUED)

Name	Function	Input Type	Output Type	tput Description				
RB3/AN9/PGM/C12IN2-	RB3	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.				
	AN9	AN	—	A/D Channel 9.				
	PGM	ST	—	Low-voltage ICSP™ Programming enable pin.				
	C12IN2-	AN	—	Comparator C1 or C2 negative input.				
RB4/AN11/P1D	RB4	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.				
	AN11	AN	—	A/D Channel 11.				
	P1D	_	CMOS	PWM output.				
RB5/AN13/T1G	RB5	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.				
	AN13	AN	—	A/D Channel 13.				
	T1G	ST	_	Timer1 Gate input.				
RB6/ICSPCLK	RB6	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.				
	ICSPCLK	ST	—	Serial Programming Clock.				
RB7/ICSPDAT	RB7	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.				
	ICSPDAT	ST	CMOS	ICSP™ Data I/O.				
RC0/T1OSO/T1CKI	RC0	ST	CMOS	General purpose I/O.				
	T1OSO	—	CMOS	Timer1 oscillator output.				
	T1CKI	ST	—	Timer1 clock input.				
RC1/T1OSI/CCP2	RC1	ST	CMOS	General purpose I/O.				
	T1OSI	ST	—	Timer1 oscillator input.				
	CCP2	ST	CMOS	Capture/Compare/PWM2.				
RC2/P1A/CCP1	RC2	ST	CMOS	General purpose I/O.				
	P1A	—	CMOS	PWM output.				
	CCP1	ST	CMOS	Capture/Compare/PWM1.				
RC3/SCK/SCL	RC3	ST	CMOS	General purpose I/O.				
	SCK	ST	CMOS	SPI clock.				
	SCL	ST	OD	I ² C™ clock.				
RC4/SDI/SDA	RC4	ST	CMOS	General purpose I/O.				
	SDI	ST	—	SPI data input.				
	SDA	ST	OD	I ² C data input/output.				
RC5/SDO	RC5	ST	CMOS	General purpose I/O.				
	SDO		CMOS	SPI data output.				
RC6/TX/CK	RC6	ST	CMOS	General purpose I/O.				
	TX		CMOS	EUSART asynchronous transmit.				
	СК	ST	CMOS	EUSART synchronous clock.				
RC7/RX/DT	RC7	ST	CMOS	General purpose I/O.				
	RX	ST	—	EUSART asynchronous input.				
	DT	ST	CMOS	EUSART synchronous data.				
RE3/MCLR/Vpp	RE3	TTL		General purpose input.				
	MCLR	ST	—	Master Clear with internal pull-up.				
N	VPP	HV	<u> </u>	Programming voltage.				
VSS	Vss	Power	<u> </u>	Ground reference.				
VDD	VDD	Power						
Legend: AN = Analog input TTL = TTL compatil HV = High Voltage	or output ble input	CMOS ST XTAL	= CMO = Schm = Crysta	S compatible input or output OD = Open Drain itt Trigger input with CMOS levels al				

Name	Function	Input Type	Output Type	Description
RA0/AN0/ULPWU/C12IN0-	RA0	TTL	CMOS	General purpose I/O.
	AN0	AN	—	A/D Channel 0 input.
	ULPWU	AN	—	Ultra Low-Power Wake-up input.
	C12IN0-	AN	—	Comparator C1 or C2 negative input.
RA1/AN1/C12IN1-	RA1	TTL	CMOS	General purpose I/O.
	AN1	AN	—	A/D Channel 1 input.
	C12IN1-	AN	—	Comparator C1 or C2 negative input.
RA2/AN2/VREF-/CVREF/C2IN+	RA2	TTL	CMOS	General purpose I/O.
	AN2	AN	—	A/D Channel 2.
	VREF-	AN	—	A/D Negative Voltage Reference input.
	CVREF	—	AN	Comparator Voltage Reference output.
	C2IN+	AN	—	Comparator C2 positive input.
RA3/AN3/VREF+/C1IN+	RA3	TTL	CMOS	General purpose I/O.
	AN3	AN	—	A/D Channel 3.
	VREF+	AN	—	A/D Positive Voltage Reference input.
	C1IN+	AN	—	Comparator C1 positive input.
RA4/T0CKI/C1OUT	RA4	TTL	CMOS	General purpose I/O.
	T0CKI	ST	—	Timer0 clock input.
	C1OUT	_	CMOS	Comparator C1 output.
RA5/AN4/SS/C2OUT	RA5	TTL	CMOS	General purpose I/O.
	AN4	AN	_	A/D Channel 4.
	SS	ST	_	Slave Select input.
	C2OUT	_	CMOS	Comparator C2 output.
RA6/OSC2/CLKOUT	RA6	TTL	CMOS	General purpose I/O.
	OSC2	_	XTAL	Crystal/Resonator.
	CLKOUT	—	CMOS	Fosc/4 output.
RA7/OSC1/CLKIN	RA7	TTL	CMOS	General purpose I/O.
	OSC1	XTAL	—	Crystal/Resonator.
	CLKIN	ST	—	External clock input/RC oscillator connection.
RB0/AN12/INT	RB0	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	AN12	AN	_	A/D Channel 12.
	INT	ST	_	External interrupt.
RB1/AN10/C12IN3-	RB1	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	AN10	AN	_	A/D Channel 10.
	C12IN3-	AN		Comparator C1 or C2 negative input.
RB2/AN8	RB2	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	AN8	AN	—	A/D Channel 8.
RB3/AN9/PGM/C12IN2-	RB3	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	AN9	AN	1 —	A/D Channel 9.
	PGM	ST	—	Low-voltage ICSP™ Programming enable pin.
	C12IN2-	AN	_	Comparator C1 or C2 negative input.
Legend: AN = Analog input TTL = TTL compati HV = High Voltage	t or output ible input	CMOS ST XTAL	= CMO = Schm = Cryst	S compatible input or output OD = Open Drain itt Trigger input with CMOS levels al

TABLE 1-2:PIC16F884/887 PINOUT DESCRIPTION

TABLE 1-2: PIC16F884/887 PINOUT DESCRIPTION (CONTINUED)

Name	Function	Input Type	Output Type	Description
RB4/AN11	RB4	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	AN11	AN	_	A/D Channel 11.
RB5/AN13/T1G	RB5	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	AN13	AN	—	A/D Channel 13.
	T1G	ST	—	Timer1 Gate input.
RB6/ICSPCLK	RB6	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	ICSPCLK	ST	—	Serial Programming Clock.
RB7/ICSPDAT	RB7	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	ICSPDAT	ST	TTL	ICSP™ Data I/O.
RC0/T1OSO/T1CKI	RC0	ST	CMOS	General purpose I/O.
	T1OSO		XTAL	Timer1 oscillator output.
	T1CKI	ST	—	Timer1 clock input.
RC1/T1OSI/CCP2	RC1	ST	CMOS	General purpose I/O.
	T1OSI	XTAL	—	Timer1 oscillator input.
	CCP2	ST	CMOS	Capture/Compare/PWM2.
RC2/P1A/CCP1	RC2	ST	CMOS	General purpose I/O.
	P1A	ST	CMOS	PWM output.
	CCP1		CMOS	Capture/Compare/PWM1.
RC3/SCK/SCL	RC3	ST	CMOS	General purpose I/O.
	SCK	ST	CMOS	SPI clock.
	SCL	ST	OD	I ² C™ clock.
RC4/SDI/SDA	RC4	ST	CMOS	General purpose I/O.
	SDI	ST	—	SPI data input.
	SDA	ST	OD	I ² C data input/output.
RC5/SDO	RC5	ST	CMOS	General purpose I/O.
	SDO	_	CMOS	SPI data output.
RC6/TX/CK	RC6	ST	CMOS	General purpose I/O.
	ТХ	—	CMOS	EUSART asynchronous transmit.
	СК	ST	CMOS	EUSART synchronous clock.
RC7/RX/DT	RC7	ST	CMOS	General purpose I/O.
	RX	ST	—	EUSART asynchronous input.
	DT	ST	CMOS	EUSART synchronous data.
RD0	RD0	TTL	CMOS	General purpose I/O.
RD1	RD1	TTL	CMOS	General purpose I/O.
RD2	RD2	TTL	CMOS	General purpose I/O.
RD3	RD3	TTL	CMOS	General purpose I/O.
RD4	RD4	TTL	CMOS	General purpose I/O.
RD5/P1B	RD5	TTL	CMOS	General purpose I/O.
	P1B	_	CMOS	PWM output.
RD6/P1C	RD6	TTL	CMOS	General purpose I/O.
	P1C	_	CMOS	PWM output.
Legend: AN = Analog inpu TTL = TTL compat HV = High Voltage	t or output ible input	CMOS ST XTAL	= CMO = Schm = Cryst	S compatible input or output OD = Open Drain itt Trigger input with CMOS levels al

PIC16F884/887 PINOUT DESCRIPTION (CONTINUED) **TABLE 1-2:**

Name	Function	Input Type	Output Type	Description
RD7/P1D	RD7	TTL	CMOS	General purpose I/O.
	P1D	AN	—	PWM output.
RE0/AN5	RE0	TTL	CMOS	General purpose I/O.
	AN5	AN	—	A/D Channel 5.
RE1/AN6	RE1	ST	CMOS	General purpose I/O.
	AN6	AN	—	A/D Channel 6.
RE2/AN7	RE2	TTL	CMOS	General purpose I/O.
	AN7	AN	—	A/D Channel 7.
RE3/MCLR/VPP	RE3	TTL	—	General purpose input.
	MCLR	ST	_	Master Clear with internal pull-up.
	Vpp	ΗV	—	Programming voltage.
Vss	Vss	Power	—	Ground reference.
VDD	Vdd	Power	—	Positive supply.
Legend: AN = Analog input	or output	CMOS	= CMO	S compatible input or output OD = Open Drain

TTL = TTL compatible input

ST = Schmitt Trigger input with CMOS levels XTAL = Crystal

HV = High Voltage

2.0 MEMORY ORGANIZATION

2.1 Program Memory Organization

The PIC16F882/883/884/886/887 has a 13-bit program counter capable of addressing a 2K x 14 (0000h-07FFh) for the PIC16F882, 4K x 14 (0000h-0FFFh) for the PIC16F883/PIC16F884, and 8K x 14 (0000h-1FFFh) for the PIC16F886/PIC16F887 program memory space. Accessing a location above these boundaries will cause a wraparound within the first 8K x 14 space. The Reset vector is at 0000h and the interrupt vector is at 0004h (see Figures 2-2 and 2-3).

FIGURE 2-1: PROGRAM MEMORY MAP AND STACK FOR THE PIC16F882



FIGURE 2-2:

PROGRAM MEMORY MAP AND STACK FOR THE PIC16F883/PIC16F884



FIGURE 2-3:

PROGRAM MEMORY MAP AND STACK FOR THE PIC16F886/PIC16F887



2.2 Data Memory Organization

The data memory (see Figures 2-2 and 2-3) is partitioned into four banks which contain the General Purpose Registers (GPR) and the Special Function Registers (SFR). The Special Function Registers are located in the first 32 locations of each bank. The General Purpose Registers, implemented as static RAM, are located in the last 96 locations of each Bank. Register locations F0h-FFh in Bank 1, 170h-17Fh in Bank 2 and 1F0h-1FFh in Bank 3, point to addresses 70h-7Fh in Bank 0. The actual number of General Purpose Resisters (GPR) implemented in each Bank depends on the device. Details are shown in Figures 2-5 and 2-6. All other RAM is unimplemented and returns '0' when read. RP<1:0> of the STATUS register are the bank select bits:

<u>RP1</u> <u>RP0</u>

- $0 \qquad 0 \qquad \rightarrow \text{Bank 0 is selected}$
- 0 1 \rightarrow Bank 1 is selected
- 1 0 \rightarrow Bank 2 is selected
- 1 1 \rightarrow Bank 3 is selected

2.2.1 GENERAL PURPOSE REGISTER FILE

The register file is organized as 128×8 in the PIC16F882, 256×8 in the PIC16F883/PIC16F884, and 368×8 in the PIC16F886/PIC16F887. Each register is accessed, either directly or indirectly, through the File Select Register (FSR) (see Section 2.4 "Indirect Addressing, INDF and FSR Registers").

Note:	The IRP and RP1 bits of the STATUS reg-								
	ister are reserved and should always be								
	maintained as '0's.								

2.2.2 SPECIAL FUNCTION REGISTERS

The Special Function Registers are registers used by the CPU and peripheral functions for controlling the desired operation of the device (see Table 2-1). These registers are static RAM.

The special registers can be classified into two sets: core and peripheral. The Special Function Registers associated with the "core" are described in this section. Those related to the operation of the peripheral features are described in the section of that peripheral feature.

FIGURE 2-4: PIC16F882 SPECIAL FUNCTION REGISTERS										
	File File File									
	Address		Address		Address		Address			
Indirect addr. ⁽¹⁾	00h	Indirect addr. ⁽¹⁾	80h	Indirect addr. (1)	100h	Indirect addr. ⁽¹⁾	180h			
TMR0	01h	OPTION_REG	81h	TMR0	101h	OPTION_REG	181h			
PCL	02h	PCL	82h	PCL	102h	PCL	182h			
STATUS	03h	STATUS	83h	STATUS	103h	STATUS	183h			
FSR	04h	FSR	84h	FSR	104h	FSR	184h			
PORTA	05h	TRISA	85h	WDTCON	105h	SRCON	185h			
PORTB	06h	TRISB	86h	PORTB	106h	TRISB	186h			
PORTC	07h	TRISC	87h	CM1CON0	107h	BAUDCTL	187h			
	08h		88h	CM2CON0	108h	ANSEL	188h			
PORTE	09h	TRISE	89h	CM2CON1	109h	ANSELH	189h			
PCLATH	0Ah	PCLATH	8Ah	PCLATH	10Ah	PCLATH	18Ah			
INTCON	0Bh	INTCON	8Bh	INTCON	10Bh	INTCON	18Bh			
PIR1	0Ch	PIE1	8Ch	EEDAT	10Ch	EECON1	18Ch			
PIR2	0Dh	PIE2	8Dh	EEADR	10Dh	EECON2 ⁽¹⁾	18Dh			
TMR1L	0Eh	PCON	8Eh	EEDATH	10Eh	Reserved	18Eh			
TMR1H	0Fh	OSCCON	8Fh	EEADRH	10Fh	Reserved	18Fh			
T1CON	10h	OSCTUNE	90h		110h		190h			
TMR2	11h	SSPCON2	91h		111h		191h			
T2CON	12h	PR2	92h		112h		192h			
SSPBUF	13h	SSPADD	93h		113h		193h			
SSPCON	14h	SSPSTAT	94h		114h		194h			
CCPR1L	15h	WPUB	95h		115h		195h			
CCPR1H	16h	IOCB	96h		116h		196h			
CCP1CON	17h	VRCON	97h		117h		197h			
RCSTA	18h	TXSTA	98h		118h		198h			
TXREG	19h	SPBRG	99h		119h		199h			
RCREG	1Ah	SPBRGH	9Ah		11Ah		19Ah			
CCPR2L	1Bh	PWM1CON	9Bh		11Bh		19Bh			
CCPR2H	1Ch	ECCPAS	9Ch		11Ch		19Ch			
CCP2CON	1Dh	PSTRCON	9Dh		11Dh		19Dh			
ADRESH	1Eh	ADRESL	9Eh		11Eh		19Eh			
ADCON0	1Fh	ADCON1	9Fh		11Fh		19Fh			
	20h	General Purpose Registers	A0h		120h		1A0h			
General		32 Butes	BFh							
Purpose		02 Dytes	COb							
Registers			0011							
96 Bytes			EEb		16Eb		1556			
		20080000	EON	20000000	170h	20000000	1505			
	756	70h-7Fh		70h-7Fh	1756	70h-7Fh	1500			
Bank O] / [1]	Book 1	F F 11	Book 2	1750	Book 2	I			
Dalik U		Dalik I		Dailk Z		Dailk 3				

FIGURE 2-5:

PIC16F883/PIC16F884 SPECIAL FUNCTION REGISTERS

	File		File		File		File
	Address		Address		Address		Address
Indirect addr. (1)	00h	Indirect addr. (1)	80h	Indirect addr. (1)	100h	Indirect addr. (1)	180h
TMR0	01h	OPTION_REG	81h	TMR0	101h	OPTION_REG	181h
PCL	02h	PCL	82h	PCL	102h	PCL	182h
STATUS	03h	STATUS	83h	STATUS	103h	STATUS	183h
FSR	04h	FSR	84h	FSR	104h	FSR	184h
PORTA	05h	TRISA	85h	WDTCON	105h	SRCON	185h
PORTB	06h	TRISB	86h	PORTB	106h	TRISB	186h
PORTC	07h	TRISC	87h	CM1CON0	107h	BAUDCTL	187h
PORTD ⁽²⁾	08h	TRISD ⁽²⁾	88h	CM2CON0	108h	ANSEL	188h
PORTE	09h	TRISE	89h	CM2CON1	109h	ANSELH	189h
PCLATH	0Ah	PCLATH	8Ah	PCLATH	10Ah	PCLATH	18Ah
INTCON	0Bh	INTCON	8Bh	INTCON	10Bh	INTCON	18Bh
PIR1	0Ch	PIE1	8Ch	EEDAT	10Ch	EECON1	18Ch
PIR2	0Dh	PIE2	8Dh	EEADR	10Dh	EECON2 ⁽¹⁾	18Dh
TMR1L	0Eh	PCON	8Eh	EEDATH	10Eh	Reserved	18Eh
TMR1H	0Fh	OSCCON	8Fh	EEADRH	10Fh	Reserved	18Fh
T1CON	10h	OSCTUNE	90h		110h		190h
TMR2	11h	SSPCON2	91h		111h		191h
T2CON	12h	PR2	92h		112h		192h
SSPBUF	13h	SSPADD	93h		113h		193h
SSPCON	14h	SSPSTAT	94h		114h		194h
CCPR1L	15h	WPUB	95h		115h		195h
CCPR1H	16h	IOCB	96h		116h		196h
CCP1CON	17h	VRCON	97h		117h		197h
RCSTA	18h	TXSTA	98h		118h		198h
TXREG	19h	SPBRG	99h		119h		199h
RCREG	1Ah	SPBRGH	9Ah		11Ah		19Ah
CCPR2L	1Bh	PWM1CON	9Bh		11Bh		19Bh
CCPR2H	1Ch	ECCPAS	9Ch		11Ch		19Ch
CCP2CON	1Dh	PSTRCON	9Dh		11Dh		19Dh
ADRESH	1Eh	ADRESL	9Eh		11Eh		19Eh
ADCON0	1Fh	ADCON1	9Fh		11Fh		19Fh
	20h	Quanta	A0h	Quanta	120h		1A0h
		General		General			
General		Registers		Registers			
Purpose				g			
Registers		80 Bytes		80 Bytes			
96 Bytes			EFh		16Fh		1EFh
,		accesses	F0h	accesses	170h	accesses	1F0h
	7Fh	70h-7Fh	FFh	70h-7Fh	17Fh	70h-7Fh	1FFh
Bank 0		Bank 1		Bank 2		Bank 3	
Unimplement	ed data me	emory locations. re	ad as 'o'.				
Note 1: Not a	ohysical re	gister.					

2: PIC16F884 only.

IGURE 2-6: PIC16F886/PIC16F887 SPECIAL FUNCTION REGISTERS										
	File		File		File					
	Address	· · · · · (1)	Address	(1)	Address	(1)	Addres			
Indirect addr. (1)	00h	Indirect addr. (1)	80h	Indirect addr. (1)	100h	Indirect addr. (1)	180h			
IMR0	01h	OPTION_REG	81h	I MR0	101h	OPTION_REG	181h			
PCL	02h	PCL	82h	PCL	102h	PCL	182h			
STATUS	03h	STATUS	83h	STATUS	103h	STATUS	183h			
FSR	04h	FSR	84h	FSR	104h	FSR	184h			
PORTA	05h	TRISA	85h	WDTCON	105h	SRCON	185h			
PORTB	06h	TRISB	86h	PORTB	106h	TRISB	186h			
PORTC	07h	TRISC	87h	CM1CON0	107h	BAUDCTL	187h			
PORTD ⁽²⁾	08h	TRISD ⁽²⁾	88h	CM2CON0	108h	ANSEL	188h			
PORTE	09h	TRISE	89h	CM2CON1	109h	ANSELH	189h			
PCLATH	0Ah	PCLATH	8Ah	PCLATH	10Ah	PCLATH	18Ah			
INTCON	0Bh	INTCON	8Bh	INTCON	10Bh	INTCON	18Bh			
PIR1	0Ch	PIE1	8Ch	EEDAT	10Ch	EECON1	18Ch			
PIR2	0Dh	PIE2	8Dh	EEADR	10Dh	EECON2 ⁽¹⁾	18Dh			
TMR1L	0Eh	PCON	8Eh	EEDATH	10Eh	Reserved	18Eh			
TMR1H	0Fh	OSCCON	8Fh	EEADRH	10Fh	Reserved	18Fh			
T1CON	10h	OSCTUNE	90h		110h		190h			
TMR2	11h	SSPCON2	91h		111h		191h			
T2CON	12h	PR2	92h		112h		192h			
SSPBUF	13h	SSPADD	93h		113h		193h			
SSPCON	14h	SSPSTAT	94h		114h		194h			
CCPR1L	15h	WPUB	95h		115h		195h			
CCPR1H	16h	IOCB	96h	General	116h	General	196h			
CCP1CON	17h	VRCON	97h	Purpose	117h	Purpose	197h			
RCSTA	18h	TXSTA	98h	Registers	118h	Registers	198h			
TXREG	19h	SPBRG	99h	16 Bytes	119h	16 Bytes	199h			
RCREG	1Ah	SPBRGH	9Ah	-	11Ah	-	19Ah			
CCPR2L	1Bh	PWM1CON	9Bh		11Bh		19Bh			
CCPR2H	1Ch	ECCPAS	9Ch		11Ch		19Ch			
CCP2CON	1Dh	PSTRCON	9Dh		11Dh		19Dh			
ADRESH	1Eh	ADRESL	9Eh		11Eh		19Eh			
ADCON0	1Fh	ADCON1	9Fh		11Fh		19Fh			
	20h		A0h		120h		1A0h			
		General		General		General				
General	3Fh	Registers		Purpose		Purpose				
Purpose	40h	. togictore		Registers		Registers				
Registers		80 Bytes		80 Bytes		80 Bytes				
96 Bytes	6Fh		EFh		16Fh		1EFh			
	70h	accesses	F0h	accesses	170h	accesses	1F0h			
	7Fh	70h-7Fh	FFh	70h-7Fh	17Fh	70h-7Fh	1FFh			
Bank 0		Bank 1	I -	Bank 2		Bank 3				
Unimplement	ed data me physical re	emory locations, re gister.	ead as '0'.							

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Page
Bank	0									_	_
00h	INDF	Addressing	this location	uses conten	ts of FSR to	address data	memory (no	ot a physical i	register)	XXXX XXXX	37,213
01h	TMR0	Timer0 Mod	dule Register	•						xxxx xxxx	73,213
02h	PCL	Program Co	ounter's (PC)) Least Signif	icant Byte					0000 0000	37,213
03h	STATUS	IRP	RP1	RP0	TO	PD	Z	DC	С	0001 1xxx	29,213
04h	FSR	Indirect Dat	a Memory A	ddress Pointe	er					xxxx xxxx	37,213
05h	PORTA ⁽³⁾	RA7	RA6	RA5	RA4	RA3	RA2	RA1	RA0	xxxx xxxx	39,213
06h	PORTB ⁽³⁾	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	xxxx xxxx	48,213
07h	PORTC ⁽³⁾	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0	xxxx xxxx	53,213
08h	PORTD ^(3,4)	RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0	xxxx xxxx	57,213
09h	PORTE ⁽³⁾	—	—	—	—	RE3	RE2 ⁽⁴⁾	RE1 ⁽⁴⁾	RE0 ⁽⁴⁾	xxxx	59,213
0Ah	PCLATH	—	—	_	Write Buffer	for upper 5 l	oits of Progra	am Counter		0 0000	37,213
0Bh	INTCON	GIE	PEIE	TOIE	INTE	RBIE	T0IF	INTF	RBIF ⁽¹⁾	0000 000x	31,213
0Ch	PIR1	_	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	-000 0000	34,213
0Dh	PIR2	OSFIF	C2IF	C1IF	EEIF	BCLIF	ULPWUIF	_	CCP2IF	0000 00-0	35,213
0Eh	TMR1L	Holding Re	gister for the	Least Signifi	cant Byte of	the 16-bit TM	R1 Register			xxxx xxxx	76,213
0Fh	TMR1H	Holding Re	Holding Register for the Most Significant Byte of the 16-bit TMR1 Register								76,213
10h	T1CON	T1GINV	TMR1GE	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N	0000 0000	79,213
11h	TMR2	Timer2 Mod	dule Register	•						0000 0000	81,213
12h	T2CON	—	TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0	-000 0000	82,213
13h	SSPBUF	Synchronou	us Serial Por	t Receive Bu	ffer/Transmit	Register				xxxx xxxx	179,213
14h	SSPCON ⁽²⁾	WCOL	SSPOV	SSPEN	СКР	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000	177,213
15h	CCPR1L	Capture/Co	mpare/PWM	Register 1 L	ow Byte (LS	3)				xxxx xxxx	126,213
16h	CCPR1H	Capture/Co	mpare/PWM	Register 1 H	ligh Byte (MS	SB)				xxxx xxxx	126,213
17h	CCP1CON	P1M1	P1M0	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	0000 0000	124,213
18h	RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000x	159,213
19h	TXREG	EUSART T	ransmit Data	Register						0000 0000	151,213
1Ah	RCREG	EUSART R	eceive Data	Register						0000 0000	156,213
1Bh	CCPR2L	Capture/Co	mpare/PWM	Register 2 L	ow Byte (LS	З)				xxxx xxxx	126,213
1Ch	CCPR2H	Capture/Co	mpare/PWM	Register 2 H	ligh Byte (MS	SB)				XXXX XXXX	126,214
1Dh	CCP2CON	—	—	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	00 0000	125,214
1Eh	ADRESH	A/D Result	Register Hig	h Byte		•		•		xxxx xxxx	99,214
1Fh	ADCON0	ADCS1	ADCS0	CHS3	CHS2	CHS1	CHS0	GO/DONE	ADON	0000 0000	104,214

PIC16F882/883/884/886/887 SPECIAL FUNCTION REGISTERS SUMMARY BANK 0 **TABLE 2-1:**

Legend: - = Unimplemented locations read as '0', u = unchanged, x = unknown, q = value depends on condition, shaded = unimplemented Note

MCLR and WDT Reset do not affect the previous value data latch. The RBIF bit will be cleared upon Reset but will set again if the 1: mismatch exists.

2: When SSPCON register bits SSPM<3:0> = 1001, any reads or writes to the SSPADD SFR address are accessed through the SSPMSK register. See Registers • and 13-4 for more detail.

Port pins with analog functions controlled by the ANSEL and ANSELH registers will read 'o' immediately after a Reset even though the 3: data latches are either undefined (POR) or unchanged (other Resets).

4: PIC16F884/PIC16F887 only.

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Page
Bank	1										
80h	INDF	Addressing	this location	uses conter	nts of FSR to	address dat	a memory (n	ot a physical	l register)	xxxx xxxx	37,213
81h	OPTION_REG	RBPU	INTEDG	TOCS	TOSE	PSA	PS2	PS1	PS0	1111 1111	30,214
82h	PCL	Program Co	ounter's (PC)	Least Signi	ficant Byte					0000 0000	37,213
83h	STATUS	IRP	RP1	RP0	TO	PD	Z	DC	С	0001 1xxx	29,213
84h	FSR	Indirect Dat	direct Data Memory Address Pointer								37,213
85h	TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	1111 1111	39,214
86h	TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	1111 1111	48,214
87h	TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	53,214
88h	TRISD ⁽³⁾	TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0	1111 1111	57,214
89h	TRISE				_	TRISE3	TRISE2 ⁽³⁾	TRISE1 ⁽³⁾	TRISE0 ⁽³⁾	1111	59,214
8Ah	PCLATH	-	-	-	Write Buffer	for the uppe	er 5 bits of th	e Program C	Counter	0 0000	37,213
8Bh	INTCON	GIE	PEIE	TOIE	INTE	RBIE	TOIF	INTF	RBIF ⁽¹⁾	0000 000x	31,213
8Ch	PIE1	_	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	-000 0000	32,214
8Dh	PIE2	OSFIE	C2IE	C1IE	EEIE	BCLIE	ULPWUIE	_	CCP2IE	0000 00-0	33,214
8Eh	PCON			ULPWUE	SBOREN	_	_	POR	BOR	01qq	36,214
8Fh	OSCCON		IRCF2	IRCF1	IRCF0	OSTS	HTS	LTS	SCS	-110 q000	62,214
90h	OSCTUNE				TUN4	TUN3	TUN2	TUN1	TUN0	0 0000	66,214
91h	SSPCON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	0000 0000	177,214
92h	PR2	Timer2 Peri	od Register							1111 1111	81,214
93h	SSPADD ⁽²⁾	Synchronou	us Serial Por	t (I ² C mode)	Address Reg	gister				0000 0000	185,214
93h	SSPMSK ⁽²⁾	MSK7	MSK6	MSK5	MSK4	MSK3	MSK2	MSK1	MSK0	1111 1111	204,214
94h	SSPSTAT	SMP	CKE	D/A	Р	S	R/W	UA	BF	0000 0000	185,214
95h	WPUB	WPUB7	WPUB6	WPUB5	WPUB4	WPUB3	WPUB2	WPUB1	WPUB0	1111 1111	49,214
96h	IOCB	IOCB7	IOCB6	IOCB5	IOCB4	IOCB3	IOCB2	IOCB1	IOCB0	0000 0000	49,214
97h	VRCON	VREN	VROE	VRR	VRSS	VR3	VR2	VR1	VR0	0000 0000	97,214
98h	TXSTA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	0000 0010	158,214
99h	SPBRG	BRG7	BRG6	BRG5	BRG4	BRG3	BRG2	BRG1	BRG0	0000 0000	161,214
9Ah	SPBRGH	BRG15	BRG14	BRG13	BRG12	BRG11	BRG10	BRG9	BRG8	0000 0000	161,214
9Bh	PWM1CON	PRSEN	PDC6	PDC5	PDC4	PDC3	PDC2	PDC1	PDC0	0000 0000	144,214
9Ch	ECCPAS	ECCPASE	ECCPAS2	ECCPAS1	ECCPAS0	PSSAC1	PSSAC0	PSSBD1	PSSBD0	0000 0000	141,214
9Dh	PSTRCON	_	_	_	STRSYNC	STRD	STRC	STRB	STRA	0 0001	145,214
9Eh	ADRESL	A/D Result	Register Lov	v Byte						xxxx xxxx	99,214
9Fh	ADCON1	ADFM	_	VCFG1	VCFG0	_	—	_	—	0-00	105,214

TARI E 2-2.	DIC16E882/883/884/886/887 S	PECIAL FUNCTION	DEGISTEDS SUMMARY	
IADLE Z-Z.	FIC10F002/003/004/000/00/ 3	FECIAL FUNCTION	REGISTERS SUMIMART	

Legend: -= Unimplemented locations read as '0', u = unchanged, x = unknown, q = value depends on condition, shaded = unimplemented Note 1: MCLR and WDT Reset do not affect the previous value data latch. The RBIF bit will be cleared upon Reset but will set again if the mismatch exists.

2: Accessible only when SSPCON register bits SSPM<3:0> = 1001.

3: PIC16F884/PIC16F887 only.

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Page
Bank	2										
100h	INDF	Addressing	this location	n uses conte	nts of FSR to a	ddress data	memory (not	a physical re	egister)	XXXX XXXX	37,213
101h	TMR0	Timer0 Mo	dule Registe		xxxx xxxx	73,213					
102h	PCL	Program C	ounter's (PC) Least Sign	ificant Byte					0000 0000	37,213
103h	STATUS	IRP	RP1	RP0	TO	PD	Z	DC	С	0001 1xxx	29,213
104h	FSR	Indirect Da	ta Memory A	ddress Poin	ter					XXXX XXXX	37,213
105h	WDTCON	—		—	WDTPS3	WDTPS2	WDTPS1	WDTPS0	SWDTEN	0 1000	221,214
106h	PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	xxxx xxxx	48,213
107h	CM1CON0	C10N	C1OUT	C10E	C1POL		C1R	C1CH1	C1CH0	0000 -000	88,214
108h	CM2CON0	C2ON	C2OUT	C2OE	C2POL	-	C2R	C2CH1	C2CH0	0000 -000	89,214
109h	CM2CON1	MC1OUT	MC2OUT	C1RSEL	C2RSEL			T1GSS	C2SYNC	000010	91,215
10Ah	PCLATH	—		_	Write Buff	er for the up	per 5 bits of t	he Program	Counter	0 0000	37,213
10Bh	INTCON	GIE	PEIE	TOIE	INTE	RBIE	T0IF	INTF	RBIF ⁽¹⁾	0000 000x	31,213
10Ch	EEDAT	EEDAT7	EEDAT6	EEDAT5	EEDAT4	EEDAT3	EEDAT2	EEDAT1	EEDAT0	0000 0000	112,215
10Dh	EEADR	EEADR7	EEADR6	EEADR5	EEADR4	EEADR3	EEADR2	EEADR1	EEADR0	0000 0000	112,215
10Eh	EEDATH	_	_	EEDATH5	EEDATH4	EEDATH3	EEDATH2	EEDATH1	EEDATH0	00 0000	112,215
10Fh	EEADRH	_	_	_	EEADRH4 ⁽²⁾	EEADRH3	EEADRH2	EEADRH1	EEADRH0	0000	112,215

TABLE 2-3: PIC16F882/883/884/886/887 SPECIAL FUNCTION REGISTERS SUMMARY BANK 2

Legend: -= Unimplemented locations read as '0', u = unchanged, x = unknown, q = value depends on condition, shaded = unimplemented Note 1: MCLR and WDT Reset does not affect the previous value data latch. The RBIF bit will be cleared upon Reset but will set again if the mismatch exists.

2: PIC16F886/PIC16F887 only.

TABLE 2-4: PIC16F882/883/884/886/887 SPECIAL FUNCTION REGISTERS SUMMARY BANK 3

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Page
Bank 3	3										
180h	INDF	Addressing	Addressing this location uses contents of FSR to address data memory (not a physical register)								37,213
181h	OPTION_REG	RBPU	INTEDG	TOCS	TOSE	PSA	PS2	PS1	PS0	1111 1111	30,214
182h	PCL	Program C	Program Counter's (PC) Least Significant Byte								37,213
183h	STATUS	IRP	RP1	RP0	TO	PD	Z	DC	С	0001 1xxx	29,213
184h	FSR	Indirect Da	ta Memory A	Address Poir	nter					xxxx xxxx	37,213
185h	SRCON	SR1	SR0	C1SEN	C2REN	PULSS	PULSR	_	FVREN	0000 00-0	93,215
186h	TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	1111 1111	48,214
187h	BAUDCTL	ABDOVF	RCIDL	_	SCKP	BRG16	_	WUE	ABDEN	01-0 0-00	160,215
188h	ANSEL	ANS7 ⁽²⁾	ANS6 ⁽²⁾	ANS5 ⁽²⁾	ANS4	ANS3	ANS2	ANS1	ANS0	1111 1111	40,215
189h	ANSELH	_	_	ANS13	ANS12	ANS11	ANS10	ANS9	ANS8	11 1111	99,215
18Ah	PCLATH	_	_	_	Write Buffe	r for the upp	er 5 bits of t	he Program	Counter	0 0000	37,213
18Bh	INTCON	GIE	PEIE	TOIE	INTE	RBIE	TOIF	INTF	RBIF ⁽¹⁾	0000 000x	31,213
18Ch	EECON1	EEPGD	—	—	—	WRERR	WREN	WR	RD	x x000	113,215
18Dh	EECON2	EEPROM (Control Regi	ster 2 (not a	physical reg	jister)					111,215

Legend: -= Unimplemented locations read as '0', u = unchanged, x = unknown, q = value depends on condition, shaded = unimplemented Note 1: MCLR and WDT Reset does not affect the previous value data latch. The RBIF bit will be cleared upon Reset but will set again if the mismatch exists.

2: PIC16F884/PIC16F887 only.

2.2.2.1 STATUS Register

The STATUS register, shown in Register 2-1, contains:

- the arithmetic status of the ALU
- the Reset status
- the bank select bits for data memory (GPR and SFR)

The STATUS register can be the destination for any instruction, like any other register. If the STATUS register is the destination for an instruction that affects the Z, DC or C bits, then the write to these three bits is disabled. These bits are set or cleared according to the device logic. Furthermore, the TO and PD bits are not writable. Therefore, the result of an instruction with the STATUS register as destination may be different than intended.

For example, CLRF STATUS, will clear the upper three bits and set the Z bit. This leaves the STATUS register as `000u uluu' (where u = unchanged).

It is recommended, therefore, that only BCF, BSF, SWAPF and MOVWF instructions are used to alter the STATUS register, because these instructions do not affect any Status bits. For other instructions not affecting any Status bits, see Section 15.0 "Instruction Set Summary"

Note 1: The <u>C and DC bits</u> operate as a Borrow and Digit Borrow out bit, respectively, in subtraction.

REGISTER 2-1: STATUS: STATUS REGISTER

R/W-0	R/W-0	R/W-0	R-1	R-1	R/W-x	R/W-x	R/W-x
IRP	RP1	RP0	TO	PD	Z	DC ⁽¹⁾	C ⁽¹⁾
bit 7							bit 0

Legend:							
R = Readable	e bit	W = Writable bit	U = Unimplemented bit,	read as '0'			
-n = Value at	POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown			
bit 7	IRP: Registe	r Bank Select bit (used f	or indirect addressing)				
	1 = Bank 2, 3 (100h-1FFh) 0 = Bank 0, 1 (00h-FFh)						
bit 6-5	RP<1:0>: Re	egister Bank Select bits (used for direct addressing)				
00 = Bank 0 (00h-7Fh)							
	01 = Bank 1	(80h-FFh) (100h-17Eh)					
	11 = Bank 3	(180h-1FFh)					
bit 4 TO: Time-out bit							
	1 = After pow 0 = A WDT ti	ver-up, CLRWDT instruction	on or SLEEP instruction				
bit 3	PD: Power-d	own bit					
	1 = After pow	ver-up or by the CLRWDT	instruction				
	0 = By execu	ition of the SLEEP instruc	ction				
bit 2	Z: Zero bit						
	1 = The resu 0 = The resu	It of an arithmetic or logi It of an arithmetic or logi	c operation is zero c operation is not zero				
bit 1	DC: Digit Ca	rry/Borrow bit (ADDWF, AI	DDLW, SUBLW, SUBWF instruction	ons) ⁽¹⁾			
	1 = A carry-o	out from the 4th low-orde	r bit of the result occurred				
	0 = No carry-	out from the 4th low-ord	er bit of the result	(4)			
bit 0	C: Carry/Bor	row bit (ADDWF, ADDLW,	SUBLW, SUBWF instructions)	(1)			
	1 = A carry-o	out from the Most Signific	cant bit of the result occurred				
Note 1: Fo	r Borrow, the po	plarity is reversed. A sub	traction is executed by adding	the two's complement of the			
se bit	cond operand. H of the source re	-or rotate (RRF, RLF) inst eaister.	ructions, this bit is loaded with	eitner the high-order or low-order			

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2.2.2.2 OPTION Register

The OPTION register, shown in Register 2-2, is a readable and writable register, which contains various control bits to configure:

- Timer0/WDT prescaler
- External INT interrupt
- Timer0
- Weak pull-ups on PORTB

Note: To achieve a 1:1 prescaler assignment for Timer0, assign the prescaler to the WDT by setting PSA bit of the OPTION register to '1'. See Section 6.3 "Timer1 Prescaler".

REGISTER 2-2: OPTION_REG: OPTION REGISTER

		—								
R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1			
RBPU	INTEDG	TOCS	TOSE	PSA	PS2	PS1	PS0			
bit 7							bit (
Legend:										
R = Readable	bit	W = Writable	e bit	U = Unimpler	mented bit, rea	d as '0'				
-n = Value at F	POR	'1' = Bit is se	et	'0' = Bit is cle	eared	x = Bit is unkr	nown			
bit 7	RBPU: PO	RTB Pull-up En	able bit							
	1 = PORTB pull-ups are disabled									
	0 = PORTB pull-ups are enabled by individual PORT latch values									
bit 6	INTEDG: Interrupt Edge Select bit									
	1 = Interrupt on rising edge of INT pin									
	0 = Interrupt on falling edge of INT pin									
bit 5	TOCS: Timer0 Clock Source Select bit									
	1 = Transition on TOCKI pin									
	0 = Internal instruction cycle clock (FOSC/4)									
bit 4	TOSE: Timer0 Source Edge Select bit									
	1 = Increment on high-to-low transition on TOCKI pin									
hit 2	0 = Increment on Iow-to-high transition on IUCKI pin									
DIL 3	PSA: Pleso	Laria assignment								
	 Prescaler is assigned to the WDI Prescaler is assigned to the Timer0 module 									
bit 2-0	PS<2:0>: F	Prescaler Rate S	Select bits							
	E	Bit Value Timer	0 Rate WDT R	ate						
	-	000 1:	2 1:1							
		001 1:	4 1:2							

1:8

1:16

1:32

1:64

1:128

1:256

010

011 100

101

110 111 1:4

1:8

1:16

1:32

1:64

1:128

2.2.2.3 INTCON Register

The INTCON register, shown in Register 2-3, is a readable and writable register, which contains the various enable and flag bits for TMR0 register overflow, PORTB change and external INT pin interrupts.

Note: Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the Global Enable bit, GIE of the INTCON register. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

REGISTER 2-3: INTCON: INTERRUPT CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-x
GIE	PEIE	TOIE	INTE	RBIE ^(1,3)	T0IF ⁽²⁾	INTF	RBIF
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	GIE: Global Interrupt Enable bit
	1 = Enables all unmasked interrupts0 = Disables all interrupts
bit 6	PEIE: Peripheral Interrupt Enable bit 1 = Enables all unmasked peripheral interrupts 0 = Disables all peripheral interrupts
bit 5	TolE: Timer0 Overflow Interrupt Enable bit 1 = Enables the Timer0 interrupt 0 = Disables the Timer0 interrupt
bit 4	INTE: INT External Interrupt Enable bit 1 = Enables the INT external interrupt 0 = Disables the INT external interrupt
bit 3	RBIE: PORTB Change Interrupt Enable bit ^(1,3) 1 = Enables the PORTB change interrupt 0 = Disables the PORTB change interrupt
bit 2	T0IF: Timer0 Overflow Interrupt Flag bit ⁽²⁾ 1 = TMR0 register has overflowed (must be cleared in software) 0 = TMR0 register did not overflow
bit 1	INTF: INT External Interrupt Flag bit 1 = The INT external interrupt occurred (must be cleared in software) 0 = The INT external interrupt did not occur
bit 0	 RBIF: PORTB Change Interrupt Flag bit 1 = When at least one of the PORTB general purpose I/O pins changed state (must be cleared in software) 0 = None of the PORTB general purpose I/O pins have changed state
Noto 1:	IOCP register must also be enchled
NULE 1:	TOE register must also be enabled.
Ζ:	TOF bit is set when timero rolls over. Timero is unchanged on Reset and should be initialized before

- 2: T0IF bit is set when Timer0 rolls over. Timer0 is unchanged on Reset and should be initialized before clearing T0IF bit.
- **3:** Includes ULPWU interrupt.

2.2.2.4 PIE1 Register

The PIE1 register contains the interrupt enable bits, as shown in Register 2-4.

Note: Bit PEIE of the INTCON register must be set to enable any peripheral interrupt.

REGISTER 2-4: PIE1: PERIPHERAL INTERRUPT ENABLE REGISTER 1

U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
_	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE
bit 7							bit 0

Legend:								
R = Readable	bit	W = Writable bit	U = Unimplemented bit, rea	ad as '0'				
-n = Value at F	POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown				
bit 7	Unimplemer	nted: Read as '0'						
bit 6	ADIE: A/D Converter (ADC) Interrupt Enable bit							
	1 = Enables 0 = Disables	the ADC interrupt the ADC interrupt						
bit 5	RCIE: EUSA	RT Receive Interrupt Enal	ble bit					
	 1 = Enables the EUSART receive interrupt 0 = Disables the EUSART receive interrupt 							
bit 4	TXIE: EUSA	RT Transmit Interrupt Ena	ble bit					
	1 = Enables 0 = Disables	the EUSART transmit inte the EUSART transmit inte	rrupt errupt					
bit 3	SSPIE: Mast	er Synchronous Serial Po	rt (MSSP) Interrupt Enable bit					
	1 = Enables 0 = Disables	the MSSP interrupt the MSSP interrupt						
bit 2	CCP1IE: CC	P1 Interrupt Enable bit						
	1 = Enables 0 = Disables	the CCP1 interrupt the CCP1 interrupt						
bit 1	TMR2IE: Tim	ner2 to PR2 Match Interrup	ot Enable bit					
	1 = Enables 0 = Disables	the Timer2 to PR2 match the Timer2 to PR2 match	interrupt interrupt					
bit 0	TMR1IE: Tim	ner1 Overflow Interrupt En	able bit					
	1 = Enables 0 = Disables	the Timer1 overflow interretter the Timer1 overflow interret	upt upt					

2.2.2.5 PIE2 Register

The PIE2 register contains the interrupt enable bits, as shown in Register 2-5.

Note: Bit PEIE of the INTCON register must be set to enable any peripheral interrupt.

REGISTER 2-5: PIE2: PERIPHERAL INTERRUPT ENABLE REGISTER 2

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	U-0	R/W-0
OSFIE	C2IE	C1IE	EEIE	BCLIE	ULPWUIE	—	CCP2IE
bit 7							bit 0

Legend:				
R = Readable	bit W = Writable	bit U = Unim	plemented bit, read	l as '0'
-n = Value at F	OR '1' = Bit is se	t '0' = Bit is	cleared	x = Bit is unknown
bit 7	OSFIE: Oscillator Fail Intern	upt Enable bit		
	1 = Enables oscillator fail in0 = Disables oscillator fail i	nterrupt nterrupt		
bit 6	C2IE: Comparator C2 Intern	upt Enable bit		
	1 = Enables Comparator C0 = Disables Comparator C	2 interrupt 2 interrupt		
bit 5	C1IE: Comparator C1 Intern	upt Enable bit		
	1 = Enables Comparator C0 = Disables Comparator C	1 interrupt 21 interrupt		
bit 4	EEIE: EEPROM Write Ope	ation Interrupt Enable bit		
	1 = Enables EEPROM writ0 = Disables EEPROM writ	e operation interrupt e operation interrupt		
bit 3	BCLIE: Bus Collision Interre	upt Enable bit		
	1 = Enables Bus Collision i0 = Disables Bus Collision	nterrupt interrupt		
bit 2	ULPWUIE: Ultra Low-Powe	r Wake-up Interrupt Enab	le bit	
	1 = Enables Ultra Low-Pow0 = Disables Ultra Low-Pow	ver Wake-up interrupt ver Wake-up interrupt		
bit 1	Unimplemented: Read as	ʻ0'		
bit 0	CCP2IE: CCP2 Interrupt Er	able bit		
	1 = Enables CCP2 interrup0 = Disables CCP2 interrup	t ot		

2.2.2.6 PIR1 Register

The PIR1 register contains the interrupt flag bits, as shown in Register 2-6.

Note: Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the Global Enable bit, GIE of the INTCON register. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

REGISTER 2-6: PIR1: PERIPHERAL INTERRUPT REQUEST REGISTER 1

U-0	R/W-0	R-0	R-0	R/W-0	R/W-0	R/W-0	R/W-0
—	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as	'0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	Unimplemented: Read as '0'
bit 6	ADIF: A/D Converter Interrupt Flag bit
	1 = A/D conversion complete (must be cleared in software)
	0 = A/D conversion has not completed or has not been started
bit 5	RCIF: EUSART Receive Interrupt Flag bit
	1 = The EUSART receive buffer is full (cleared by reading RCREG)
	0 = The EUSART receive buffer is not full
bit 4	TXIF: EUSART Transmit Interrupt Flag bit
	1 = The EUSART transmit buffer is empty (cleared by writing to TXREG)
L'I O	0 = The EUSART transmit burrer is full
bit 3	SSPIF: Master Synchronous Serial Port (MSSP) Interrupt Flag bit
	1 = The MSSP interrupt condition has occurred, and must be cleared in software before returning from the Interrupt Service Routine. The conditions that will set this bit are: SDI
	A transmission/reception has taken place
	I ² C Slave/Master
	A transmission/reception has taken place
	I ² C Master
	The initiated Staft condition was completed by the MSSP module
	The initiated restart condition was completed by the MSST module
	The initiated Acknowledge condition was completed by the MSSP module
	A Start condition occurred while the MSSP module was idle (Multi-master system)
	A Stop condition occurred while the MSSP module was idle (Multi-master system)
	0 = No MSSP interrupt condition has occurred
bit 2	CCP1IF: CCP1 Interrupt Flag bit
	<u>Capture mode:</u>
	1 = A TMRT register capture occurred (must be cleared in software) 0 = No TMR1 register capture occurred
	Compare mode:
	1 = A TMR1 register compare match occurred (must be cleared in software)
	0 = No TMR1 register compare match occurred
	<u>PWM mode</u> :
1. 1. A	THREE Traces to BRO between the bit
bit 1	IMR2IF: Timer2 to PR2 Interrupt Flag bit
	1 = A timer2 to PR2 match occurred (must be cleared in software) 0 = No Timer2 to PR2 match occurred
bit 0	TMR1IE: Timer1 Overflow Interrupt Eleg hit
DIL U	1 = The TMR1 register overflowed (must be cleared in software)
	0 = The TMR1 register did not overflow
	-
2.2.2.7 PIR2 Register

The PIR2 register contains the interrupt flag bits, as shown in Register 2-7.

Note: Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the Global Enable bit, GIE of the INTCON register. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

REGISTER 2-7: PIR2: PERIPHERAL INTERRUPT REQUEST REGISTER 2

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	U-0	R/W-0
OSFIF	C2IF	C1IF	EEIF	BCLIF	ULPWUIF	—	CCP2IF
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	OSFIF: Oscillator Fail Interrupt Flag bit
	 1 = System oscillator failed, clock input has changed to INTOSC (must be cleared in software) 0 = System clock operating
bit 6	C2IF: Comparator C2 Interrupt Flag bit
	 1 = Comparator output (C2OUT bit) has changed (must be cleared in software) 0 = Comparator output (C2OUT bit) has not changed
bit 5	C1IF: Comparator C1 Interrupt Flag bit
	 1 = Comparator output (C1OUT bit) has changed (must be cleared in software) 0 = Comparator output (C1OUT bit) has not changed
bit 4	EEIF: EE Write Operation Interrupt Flag bit
	1 = Write operation completed (must be cleared in software)0 = Write operation has not completed or has not started
bit 3	BCLIF: Bus Collision Interrupt Flag bit
	 1 = A bus collision has occurred in the MSSP when configured for I²C Master mode 0 = No bus collision has occurred
bit 2	ULPWUIF: Ultra Low-Power Wake-up Interrupt Flag bit
	 1 = Wake-up condition has occurred (must be cleared in software) 0 = No Wake-up condition has occurred
bit 1	Unimplemented: Read as '0'
bit 0	CCP2IF: CCP2 Interrupt Flag bit
	Capture mode:
	 1 = A TMR1 register capture occurred (must be cleared in software) 0 = No TMR1 register capture occurred
	Compare mode:
	1 = A TMR1 register compare match occurred (must be cleared in software)
	0 = NO IMR1 register compare match occurred PWM mode:
	Unused in this mode

2.2.2.8 PCON Register

The Power Control (PCON) register (see Register 2-8) contains flag bits to differentiate between a:

- Power-on Reset (POR)
- Brown-out Reset (BOR)
- Watchdog Timer Reset (WDT)
- External MCLR Reset

The PCON register also controls the Ultra Low-Power Wake-up and software enable of the BOR.

REGISTER 2-8: PCON: POWER CONTROL REGISTER

U-0	U-0	R/W-0	R/W-1	U-0	U-0	R/W-0	R/W-x
—	—	ULPWUE	SBOREN ⁽¹⁾	—	—	POR	BOR
bit 7							bit 0

Legend:							
R = Readable bit W		W = Writable bit	U = Unimplemented bit,	read as '0'			
-n = Value a	at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown			
bit 7-6	Unimpleme	nted: Read as '0'					
bit 5	ULPWUE: L	Jltra Low-Power Wake-up	o Enable bit				
	1 = Ultra Lov 0 = Ultra Lov	w-Power Wake-up enable w-Power Wake-up disabl	ed ed				
bit 4	SBOREN: S	Software BOR Enable bit	1)				
	1 = BOR en 0 = BOR dis	abled abled					
bit 3-2	Unimpleme	Unimplemented: Read as '0'					
bit 1	POR: Power-on Reset Status bit						
 1 = No Power-on Reset occurred 0 = A Power-on Reset occurred (must be set in software after a Power-on Reset of the set in software after a Power-on Reset of the set of the set in software after a Power-on Reset of the set of t							
bit 0	BOR: Brown	n-out Reset Status bit					
	1 = No Brow 0 = A Brown	vn-out Reset occurred n-out Reset occurred (mu	st be set in software after a Br	own-out Reset occurs)			
	1 = No Brow 0 = A Brown	vn-out Reset occurred n-out Reset occurred (mu	ist be set in software after a Br	own-out Reset occurs)			

Note 1: BOREN<1:0> = 01 in the Configuration Word Register 1 for this bit to control the \overline{BOR} .

2.3 PCL and PCLATH

The Program Counter (PC) is 13 bits wide. The low byte comes from the PCL register, which is a readable and writable register. The high byte (PC<12:8>) is not directly readable or writable and comes from PCLATH. On any Reset, the PC is cleared. Figure 2-7 shows the two situations for the loading of the PC. The upper example in Figure 2-7 shows how the PC is loaded on a write to PCL (PCLATH<4:0> \rightarrow PCH). The lower example in Figure 2-7 shows how the PC is loaded during a CALL or GOTO instruction (PCLATH<4:3> \rightarrow PCH).

FIGURE 2-7: LOADING OF PC IN DIFFERENT SITUATIONS



2.3.1 MODIFYING PCL

Executing any instruction with the PCL register as the destination simultaneously causes the Program Counter PC<12:8> bits (PCH) to be replaced by the contents of the PCLATH register. This allows the entire contents of the program counter to be changed by writing the desired upper 5 bits to the PCLATH register. When the lower 8 bits are written to the PCL register, all 13 bits of the program counter will change to the values contained in the PCLATH register and those being written to the PCL register.

A computed GOTO is accomplished by adding an offset to the program counter (ADDWF PCL). Care should be exercised when jumping into a look-up table or program branch table (computed GOTO) by modifying the PCL register. Assuming that PCLATH is set to the table start address, if the table length is greater than 255 instructions or if the lower 8 bits of the memory address rolls over from 0xFF to 0x00 in the middle of the table, then PCLATH must be incremented for each address rollover that occurs between the table beginning and the target location within the table.

For more information refer to Application Note AN556, *"Implementing a Table Read"* (DS00556).

2.3.2 STACK

The PIC16F882/883/884/886/887 devices have an 8-level x 13-bit wide hardware stack (see Figures 2-2 and 2-3). The stack space is not part of either program or data space and the Stack Pointer is not readable or writable. The PC is PUSHed onto the stack when a CALL instruction is executed or an interrupt causes a branch. The stack is POPed in the event of a RETURN, RETLW or a RETFIE instruction execution. PCLATH is not affected by a PUSH or POP operation.

The stack operates as a circular buffer. This means that after the stack has been PUSHed eight times, the ninth push overwrites the value that was stored from the first push. The tenth push overwrites the second push (and so on).

Note 1:	There are no Status bits to indicate stack overflow or stack underflow conditions.						
2:	There are no instructions/mnemonics called PUSH or POP. These are actions that occur from the execution of the CALL, RETURN, RETLW and RETFIE instructions or the vectoring to an interrupt address.						

2.4 Indirect Addressing, INDF and FSR Registers

The INDF register is not a physical register. Addressing the INDF register will cause indirect addressing.

Indirect addressing is possible by using the INDF register. Any instruction using the INDF register actually accesses data pointed to by the File Select Register (FSR). Reading INDF itself indirectly will produce 00h. Writing to the INDF register indirectly results in a no operation (although Status bits may be affected). An effective 9-bit address is obtained by concatenating the 8-bit FSR and the IRP bit of the STATUS register, as shown in Figure 2-8.

A simple program to clear RAM location 20h-2Fh using indirect addressing is shown in Example 2-1.

EXAM	PLE 2-1:	INDIRECT ADDRESSING			
NEXT	MOVLW MOVWF CLRF INCF BTFSS GOTO	0x20 FSR INDF FSR FSR,4 NEXT	;initialize pointer ;to RAM ;clear INDF register ;inc pointer ;all done? ;no clear next		
CONTINUE			;yes continue		



FIGURE 2-8: DIRECT/INDIRECT ADDRESSING PIC16F882/883/884/886/887

3.0 I/O PORTS

There are as many as thirty-five general purpose I/O pins available. Depending on which peripherals are enabled, some or all of the pins may not be available as general purpose I/O. In general, when a peripheral is enabled, the associated pin may not be used as a general purpose I/O pin.

3.1 PORTA and the TRISA Registers

PORTA is a 8-bit wide, bidirectional port. The corresponding data direction register is TRISA (Register 3-2). Setting a TRISA bit (= 1) will make the corresponding PORTA pin an input (i.e., disable the output driver). Clearing a TRISA bit (= 0) will make the corresponding PORTA pin an output (i.e., enables output driver and puts the contents of the output latch on the selected pin). Example 3-1 shows how to initialize PORTA.

Reading the PORTA register (Register 3-1) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write

REGISTER 3-1: PORTA: PORTA REGISTER

operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch.

The TRISA register (Register 3-2) controls the PORTA pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISA register are maintained set when using them as analog inputs. I/O pins configured as analog input always read '0'.

Note:	The ANSEL register must be initialized to						
	configure an analog channel as a digital						
	input. Pins configured as analog inputs wil						
	read '0'.						

BANKSEL	PORTA	;
CLRF	PORTA	;Init PORTA
BANKSEL	ANSEL	;
CLRF	ANSEL	;digital I/O
BCF	STATUS, RP1	;Bank 1
BANKSEL	TRISA	;
MOVLW	0Ch	;Set RA<3:2> as inputs
MOVWF	TRISA	;and set RA<5:4,1:0>
		;as outputs

| R/W-x |
|-------|-------|-------|-------|-------|-------|-------|-------|
| RA7 | RA6 | RA5 | RA4 | RA3 | RA2 | RA1 | RA0 |
| bit 7 | | | | | | | bit 0 |

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as	'0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **RA<7:0>:** PORTA I/O Pin bit 1 = Port pin is > VIH

0 = Port pin is < VIL

REGISTER 3-2: TRISA: PORTA TRI-STATE REGISTER

R/W-1 ⁽¹⁾	R/W-1 ⁽¹⁾	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0
bit 7							bit 0

W = Writable bit	U = Unimplemented bit, read as	'0'
'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown
	W = Writable bit '1' = Bit is set	W = Writable bitU = Unimplemented bit, read as'1' = Bit is set'0' = Bit is cleared

bit 7-0

TRISA<7:0>: PORTA Tri-State Control bit 1 = PORTA pin configured as an input (tri-stated)

0 = PORTA pin configured as an input (in s

Note 1: TRISA<7:6> always reads '1' in XT, HS and LP Oscillator modes.

3.2 Additional Pin Functions

RA0 also has an Ultra Low-Power Wake-up option. The next three sections describe these functions.

3.2.1 ANSEL REGISTER

The ANSEL register (Register 3-3) is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSEL bit high will cause all digital reads on the pin to be read as '0' and allow analog functions on the pin to operate correctly.

The state of the ANSEL bits has no affect on digital output functions. A pin with TRIS clear and ANSEL set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

REGISTER 3-3: ANSEL: ANALOG SELECT REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
ANS7 ⁽²⁾	ANS6 ⁽²⁾	ANS5 ⁽²⁾	ANS4	ANS3	ANS2	ANS1	ANS0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 ANS<7:0>: Analog Select bits

Γ.

Analog select between analog or digital function on pins AN<7:0>, respectively.

- 1 =Analog input. Pin is assigned as analog input⁽¹⁾.
- 0 = Digital I/O. Pin is assigned to port or special function.
- **Note 1:** Setting a pin to an analog input automatically disables the digital input circuitry, weak pull-ups, and interrupt-on-change if available. The corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.
 - 2: Not implemented on PIC16F883/886.

3.2.2 ULTRA LOW-POWER WAKE-UP

The Ultra Low-Power Wake-up (ULPWU) on RA0 allows a slow falling voltage to generate an interrupt-on-change on RA0 without excess current consumption. The mode is selected by setting the ULPWUE bit of the PCON register. This enables a small current sink, which can be used to discharge a capacitor on RA0.

Follow these steps to use this feature:

- a) Charge the capacitor on RA0 by configuring the RA0 pin to output (= 1).
- b) Configure RA0 as an input.
- c) Enable interrupt-on-change for RA0.
- d) Set the ULPWUE bit of the PCON register to begin the capacitor discharge.
- e) Execute a **SLEEP** instruction.

When the voltage on RA0 drops below VIL, an interrupt will be generated which will cause the device to wake-up and execute the next instruction. If the GIE bit of the INTCON register is set, the device will then call the interrupt vector (0004h). See **Section 3.4.3** "Interrupt-on-Change" for more information.

This feature provides a low-power technique for periodically waking up the device from Sleep. The time-out is dependent on the discharge time of the RC circuit on RA0. See Example 3-2 for initializing the Ultra Low-Power Wake-up module. A series resistor between RA0 and the external capacitor provides overcurrent protection for the RA0/AN0/ULPWU/C12IN0- pin and can allow for software calibration of the time-out (see Figure 3-1). A timer can be used to measure the charge time and discharge time of the capacitor. The charge time can then be adjusted to provide the desired interrupt delay. This technique will compensate for the affects of temperature, voltage and component accuracy. The Ultra Low-Power Wake-up peripheral can also be configured as a simple Programmable Low Voltage Detect or temperature sensor.

Note: For more information, refer to AN879, "Using the Microchip Ultra Low-Power Wake-up Module" Application Note (DS00879).

EXAMPLE 3-2: ULTRA LOW-POWER WAKE-UP INITIALIZATION

BANKSEL	PORTA	;
BSF	PORTA,0	;Set RA0 data latch
BANKSEL	ANSEL	;
BCF	ANSEL,0	;RA0 to digital I/O
BANKSEL	TRISA	;
BCF	TRISA,0	;Output high to
CALL	CapDelay	;charge capacitor
BANKSEL	PIR2	;
BCF	PIR2,ULPWUIF	;Clear flag
BSF	PCON,ULPWUE	;Enable ULP Wake-up
		· · · · · · · · · · · · · · · · · · ·
BSF	IOCB,0	;Select RA0 IOC
BSF BSF	IOCB,0 TRISA,0	;Select RA0 IOC ;RA0 to input
BSF BSF MOVLW	IOCB,0 TRISA,0 B'10001000'	;Select RA0 IOC ;RA0 to input ;Enable interrupt
BSF BSF MOVLW MOVWF	IOCB,0 TRISA,0 B'10001000' INTCON	;Select RA0 IOC ;RA0 to input ;Enable interrupt ;and clear flag
BSF BSF MOVLW MOVWF SLEEP	IOCB,0 TRISA,0 B'10001000' INTCON	;Select RA0 IOC ;RA0 to input ;Enable interrupt ;and clear flag ;Wait for IOC
BSF BSF MOVLW MOVWF SLEEP NOP	IOCB,0 TRISA,0 B'10001000' INTCON	;Select RA0 IOC ;RA0 to input ;Enable interrupt ;and clear flag ;Wait for IOC ;

3.2.3 PIN DESCRIPTIONS AND DIAGRAMS

Each PORTA pin is multiplexed with other functions. The pins and their combined functions are briefly described here. For specific information about individual functions such as the comparator or the A/D Converter (ADC), refer to the appropriate section in this data sheet.

3.2.3.1 RA0/AN0/ULPWU/C12IN0-

Figure 3-1 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC
- a negative analog input to Comparator C1 or C2
- an analog input for the Ultra Low-Power Wake-up



FIGURE 3-1: BLOCK DIAGRAM OF RA0

3.2.3.2 RA1/AN1/C12IN1-

Figure 3-2 shows the diagram for this pin. This pin is configurable to function as one of the following:

• a general purpose I/O

- an analog input for the ADC
- a negative analog input to Comparator C1 or C2

FIGURE	3-2: BLOCK DIAGRAIN OF F	KA 1
Data Bus Data Bus WR PORTA		Pin
WR TRISA RD TRISA	Analog ⁽¹⁾	
RD PORTA		
•	To Comparator	
4	To A/D Converter	
Note 1	: ANSEL determines Analog Input mode.	

3.2.3.3 RA2/AN2/VREF-/CVREF/C2IN+

Figure 3-3 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC
- a negative voltage reference input for the ADC and CVREF
- a comparator voltage reference output
- a positive analog input to Comparator C2

FIGURE 3-3: BLOCK DIAGRAM OF RA2



PIC16F882/883/884/886/887

3.2.3.4 RA3/AN3/VREF+/C1IN+

Figure 3-4 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose input
- an analog input for the ADC
- a positive voltage reference input for the ADC and CVREF
- a positive analog input to Comparator C1



3.2.3.5 RA4/T0CKI/C1OUT

Figure 3-5 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- a clock input for Timer0
- a digital output from Comparator C1





3.2.3.6 RA5/AN4/SS/C2OUT

Figure 3-6 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC
- a slave select input
- a digital output from Comparator C2



3.2.3.7 RA6/OSC2/CLKOUT

Figure 3-7 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- a crystal/resonator connection
- a clock output



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3.2.3.8 RA7/OSC1/CLKIN

Figure 3-8 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- a crystal/resonator connection
- a clock input



TABLE 3-1: SUMMARY OF REGISTERS ASSOCIATED WITH PORTA

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ADCON0	ADCS1	ADCS0	CHS3	CHS2	CHS1	CHS0	GO/DONE	ADON	0000 0000	0000 0000
ANSEL	ANS7	ANS6	ANS5	ANS4	ANS3	ANS2	ANS1	ANS0	1111 1111	1111 1111
CM1CON0	C10N	C1OUT	C1OE	C1POL	—	C1R	C1CH1	C1CH0	0000 -000	0000 -000
CM2CON0	C2ON	C2OUT	C2OE	C2POL	_	C2R	C2CH1	C2CH0	0000 -000	0000 -000
CM2CON1	MC1OUT	MC2OUT	C1RSEL	C2RSEL	_	_	T1GSS	C2SYNC	000010	000010
PCON	—	—	ULPWUE	SBOREN	—	—	POR	BOR	01qq	0uuu
OPTION_REG	RBPU	INTEDG	TOCS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
PORTA	RA7	RA6	RA5	RA4	RA3	RA2	RA1	RA0	xxxx xxxx	uuuu uuuu
SSPCON	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000	0000 0000
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	1111 1111	1111 1111

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by PORTA.

3.3 PORTB and TRISB Registers

PORTB is an 8-bit wide, bidirectional port. The corresponding data direction register is TRISB (Register 3-6). Setting a TRISB bit (= 1) will make the corresponding PORTB pin an input (i.e., put the corresponding output driver in a High-Impedance mode). Clearing a TRISB bit (= 0) will make the corresponding PORTB pin an output (i.e., enable the output driver and put the contents of the output latch on the selected pin). Example 3-3 shows how to initialize PORTB.

Reading the PORTB register (Register 3-5) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch.

The TRISB register (Register 3-6) controls the PORTB pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISB register are maintained set when using them as analog inputs. I/O pins configured as analog input always read '0'. Example 3-3 shows how to initialize PORTB.

EXAMPLE 3-3: INITIALIZING PORTB

BANKSEL	J PORTB	;
CLRF	PORTB	;Init PORTB
BANKSEI	TRISB	;
MOVLW	B'11110000'	;Set RB<7:4> as inputs
		;and RB<3:0> as outputs
MOVWF	TRISB	;

Note: The ANSELH register must be initialized to configure an analog channel as a digital input. Pins configured as analog inputs will read '0'.

3.4 Additional PORTB Pin Functions

PORTB pins RB<7:0> on the device family device have an interrupt-on-change option and a weak pull-up option. The following three sections describe these PORTB pin functions.

Every PORTB pin on this device family has an interrupt-on-change option and a weak pull-up option.

3.4.1 ANSELH REGISTER

The ANSELH register (Register 3-4) is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSELH bit high will cause all digital reads on the pin to be read as '0' and allow analog functions on the pin to operate correctly.

The state of the ANSELH bits has no affect on digital output functions. A pin with TRIS clear and ANSELH set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

3.4.2 WEAK PULL-UPS

Each of the PORTB pins has an individually configurable internal weak pull-up. Control bits WPUB<7:0> enable or disable each pull-up (see Register 3-7). Each weak pull-up is automatically turned off when the port pin is configured as an output. All pull-ups are disabled on a Power-on Reset by the RBPU bit of the OPTION register.

3.4.3 INTERRUPT-ON-CHANGE

All of the PORTB pins are individually configurable as an interrupt-on-change pin. Control bits IOCB<7:0> enable or disable the interrupt function for each pin. Refer to Register 3-8. The interrupt-on-change feature is disabled on a Power-on Reset.

For enabled interrupt-on-change pins, the present value is compared with the old value latched on the last read of PORTB to determine which bits have changed or mismatched the old value. The 'mismatch' outputs of the last read are OR'd together to set the PORTB Change Interrupt flag bit (RBIF) in the INTCON register.

This interrupt can wake the device from Sleep. The user, in the Interrupt Service Routine, clears the interrupt by:

- a) Any read or write of PORTB. This will end the mismatch condition.
- b) Clear the flag bit RBIF.

A mismatch condition will continue to set flag bit RBIF. Reading or writing PORTB will end the mismatch condition and allow flag bit RBIF to be cleared. The latch holding the last read value is not affected by a MCLR nor Brown-out Reset. After these Resets, the RBIF flag will continue to be set if a mismatch is present.

Note: If a change on the I/O pin should occur when the read operation is being executed (start of the Q2 cycle), then the RBIF interrupt flag may not get set. Furthermore, since a read or write on a port affects all bits of that port, care must be taken when using multiple pins in Interrupt-on-Change mode. Changes on one pin may not be seen while servicing changes on another pin.

U-0	U-0	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
—		ANS13	ANS12	ANS11	ANS10	ANS9	ANS8
bit 7							bit 0
Legend:							
R = Readable I	bit	W = Writable bit U = Unimplemented bit, read as '0'					
-n = Value at P	OR	'1' = Bit is set		'0' = Bit is cle	0' = Bit is cleared x = Bit is unknown		

REGISTER 3-4: ANSELH: ANALOG SELECT HIGH REGISTER

bit 7-6 Unimplemented: Read as '0'

bit 5-0 ANS<13:8>: Analog Select bits Analog select between analog or digital function on pins AN<13:8>, respectively. 1 = Analog input. Pin is assigned as analog input⁽¹⁾. 0 = Digital I/O. Pin is assigned to port or special function.

Note 1: Setting a pin to an analog input automatically disables the digital input circuitry, weak pull-ups, and interrupt-on-change if available. The corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.

REGISTER 3-5: PORTB: PORTB REGISTER

| R/W-x |
|-------|-------|-------|-------|-------|-------|-------|-------|
| RB7 | RB6 | RB5 | RB4 | RB3 | RB2 | RB1 | RB0 |
| bit 7 | | | | | | | bit 0 |

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **RB<7:0>**: PORTB I/O Pin bit 1 = Port pin is > VIH 0 = Port pin is < VIL

REGISTER 3-6: TRISB: PORTB TRI-STATE REGISTER

| R/W-1 |
|--------|--------|--------|--------|--------|--------|--------|--------|
| TRISB7 | TRISB6 | TRISB5 | TRISB4 | TRISB3 | TRISB2 | TRISB1 | TRISB0 |
| bit 7 | | | | | | | bit 0 |

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	l as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 TRISB<7:0>: PORTB Tri-State Control bit

1 = PORTB pin configured as an input (tri-stated)

0 = PORTB pin configured as an output

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
WPUB7	WPUB6	WPUB5	WPUB4	WPUB3	WPUB2	WPUB1	WPUB0
bit 7							bit 0
Legend:							
R = Readable	bit	W = Writable	bit	it U = Unimplemented bit, read as '0'			
-n = Value at P	OR	'1' = Bit is set		'0' = Bit is cleared x = Bit is unkno			nown

REGISTER 3-7: WPUB: WEAK PULL-UP PORTB REGISTER

bit 7-0 WPUB<7:0>: Weak Pull-up Register bit

- 1 = Pull-up enabled
- 0 = Pull-up disabled

Note 1: Global RBPU bit of the OPTION register must be cleared for individual pull-ups to be enabled.

2: The weak pull-up device is automatically disabled if the pin is in configured as an output.

REGISTER 3-8: IOCB: INTERRUPT-ON-CHANGE PORTB REGISTER

| R/W-0 |
|-------|-------|-------|-------|-------|-------|-------|-------|
| IOCB7 | IOCB6 | IOCB5 | IOCB4 | IOCB3 | IOCB2 | IOCB1 | IOCB0 |
| bit 7 | | | | | | | bit 0 |

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **IOCB<7:0>:** Interrupt-on-Change PORTB Control bit

- 1 = Interrupt-on-change enabled
- 0 = Interrupt-on-change disabled

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3.4.4 PIN DESCRIPTIONS AND DIAGRAMS

Each PORTB pin is multiplexed with other functions. The pins and their combined functions are briefly described here. For specific information about individual functions such as the SSP, I²C or interrupts, refer to the appropriate section in this data sheet.

3.4.4.1 RB0/AN12/INT

Figure 3-9 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC
- an external edge triggered interrupt

3.4.4.2 RB1/AN10/P1C⁽¹⁾/C12IN3-

Figure 3-9 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC
- a PWM output⁽¹⁾
- an analog input to Comparator C1 or C2

Note 1: P1C is available on PIC16F882/883/886 only.

3.4.4.3 RB2/AN8/P1B⁽¹⁾

Figure 3-9 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC
- a PWM output⁽¹⁾

Note 1: P1B is available on PIC16F882/883/886 only.

3.4.4.4 RB3/AN9/PGM/C12IN2-

Figure 3-9 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC
- Low-voltage In-Circuit Serial Programming enable
 pin
- an analog input to Comparator C1 or C2

FIGURE 3-9: BL

BLOCK DIAGRAM OF RB<3:0>



3.4.4.5 RB4/AN11/P1D⁽¹⁾

Figure 3-10 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC
- a PWM output⁽¹⁾

Note 1: P1D is available on PIC16F882/883/886 only.

3.4.4.6 RB5/AN13/T1G

Figure 3-10 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC
- a Timer1 gate input

FIGURE 3-10: BLOCK DIAGRAM OF RB<7:4>

3.4.4.7 RB6/ICSPCLK

Figure 3-10 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- In-Circuit Serial Programming clock

3.4.4.8 RB7/ICSPDAT

Figure 3-10 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- In-Circuit Serial Programming data



Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ANSELH	—	—	ANS13	ANS12	ANS11	ANS10	ANS9	ANS8	11 1111	11 1111
CCP1CON	P1M1	P1M0	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	0000 0000	0000 0000
CM2CON1	MC1OUT	MC2OUT	C1RSEL	C2RSEL		_	T1GSS	C2SYNC	000010	000010
IOCB	IOCB7	IOCB6	IOCB5	IOCB4	IOCB3	IOCB2	IOCB1	IOCB0	0000 0000	0000 0000
INTCON	GIE	PEIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	0000 000x
OPTION_REG	RBPU	INTEDG	TOCS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	xxxx xxxx	uuuu uuuu
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	1111 1111	1111 1111
WPUB	WPUB7	WPUB6	WPUB5	WPUB4	WPUB3	WPUB2	WPUB1	WPUB0	1111 1111	1111 1111

TABLE 3-2: SUMMARY OF REGISTERS ASSOCIATED WITH PORTB

Legend: x = unknown, u = unchanged, - = unimplemented read as '0'. Shaded cells are not used by PORTB.

3.5 **PORTC and TRISC Registers**

PORTC is a 8-bit wide, bidirectional port. The corresponding data direction register is TRISC (Register 3-10). Setting a TRISC bit (= 1) will make the corresponding PORTC pin an input (i.e., put the corresponding output driver in a High-Impedance mode). Clearing a TRISC bit (= 0) will make the corresponding PORTC pin an output (i.e., enable the output driver and put the contents of the output latch on the selected pin). Example 3-4 shows how to initialize PORTC.

Reading the PORTC register (Register 3-9) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch.

The TRISC register (Register 3-10) controls the PORTC pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISC register are maintained set when using them as analog inputs. I/O pins configured as analog input always read '0'.

EXAMPLE 3-4: INITIALIZING PORTC BANKSEL PORTC ; CLRF PORTC ; Init PORTC BANKSEL TRISC ; MOVLW B'00001100' ;Set RC<3:2> as inputs MOVWF TRISC ; and set RC<7:4,1:0> ;as outputs

REGISTER 3-9: PORTC: PORTC REGISTER

| R/W-x |
|-------|-------|-------|-------|-------|-------|-------|-------|
| RC7 | RC6 | RC5 | RC4 | RC3 | RC2 | RC1 | RC0 |
| bit 7 | | | | | | | bit 0 |

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	l as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **RC<7:0>**: PORTC General Purpose I/O Pin bit 1 = Port pin is > VIH 0 = Port pin is < VIL

REGISTER 3-10: TRISC: PORTC TRI-STATE REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1 ⁽¹⁾	R/W-1 ⁽¹⁾
TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 TRISC<7:0>: PORTC Tri-State Control bit

1 = PORTC pin configured as an input (tri-stated)

0 = PORTC pin configured as an output

Note 1: TRISC<1:0> always reads '1' in LP Oscillator mode.

3.5.1 RC0/T1OSO/T1CKI

Figure 3-11 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- a Timer1 oscillator output
- a Timer1 clock input



3.5.2 RC1/T1OSI/CCP2

Figure 3-12 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- a Timer1 oscillator input
- a Capture input and Compare/PWM output for Comparator C2





3.5.3 RC2/P1A/CCP1

Figure 3-13 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- a PWM output
- a Capture input and Compare output for Comparator C1





3.5.4 RC3/SCK/SCL

Figure 3-14 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- · a SPI clock
- an I²C[™] clock

FIGURE 3-14: BLOCK DIAGRAM OF RC3



3.5.5 RC4/SDI/SDA

Figure 3-15 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- a SPI data I/O
- an I²C data I/O

FIGURE 3-15: BLOCK DIAGRAM OF RC4



3.5.6 RC5/SDO

Figure 3-16 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- · a serial data output

FIGURE 3-16:



3.5.7 RC6/TX/CK

Figure 3-17 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- · an asynchronous serial output
- a synchronous clock I/O



RC7/RX/DT 3.5.8

Figure 3-18 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- · an asynchronous serial input
- a synchronous serial data I/O







TABLE 3-3: SUMMARY OF REGISTERS ASSOCIATED WITH PORTC

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
CCP1CON	P1M1	P1M0	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	0000 0000	0000 0000
CCP2CON	—	_	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	00 0000	00 0000
PORTC	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0	xxxx xxxx	uuuu uuuu
PSTRCON	_	_	_	STRSYNC	STRD	STRC	STRB	STRA	0 0001	0 0001
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000x	0000 000x
SSPCON	WCOL	SSPOV	SSPEN	СКР	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000	0000 0000
T1CON	T1GINV	TMR1GE	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR1ON	0000 0000	0000 0000
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by PORTC.

3.6 **PORTD and TRISD Registers**

PORTD⁽¹⁾ is a 8-bit wide, bidirectional port. The corresponding data direction register is TRISD (Register 3-12). Setting a TRISD bit (= 1) will make the corresponding PORTD pin an input (i.e., put the corresponding output driver in a High-Impedance mode). Clearing a TRISD bit (= 0) will make the corresponding PORTD pin an output (i.e., enable the output driver and put the contents of the output latch on the selected pin). Example 3-5 shows how to initialize PORTD.

Reading the PORTD register (Register 3-11) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch.

Note 1: PORTD is available on PIC16F884/887 only.

PORTD: PORTD REGISTER **REGISTER 3-11:**

R/W-x R/W-x R/W-x R/W-x R/W-x R/W-x R/W-x R/W-x RD7 RD6 RD5 RD4 RD3 RD2 RD1 RD0 bit 7 bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	l as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 RD<7:0>: PORTD General Purpose I/O Pin bit

1 = Port pin is > VIH

0 = Port pin is < VIL

REGISTER 3-12: TRISD: PORTD TRI-STATE REGISTER

| R/W-1 |
|--------|--------|--------|--------|--------|--------|--------|--------|
| TRISD7 | TRISD6 | TRISD5 | TRISD4 | TRISD3 | TRISD2 | TRISD1 | TRISD0 |
| bit 7 | | | | | | | bit 0 |

Legend:				
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown	

bit 7-0 TRISD<7:0>: PORTD Tri-State Control bit 1 = PORTD pin configured as an input (tri-stated) 0 = PORTD pin configured as an output

The TRISD register (Register 3-12) controls the PORTD pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISD register are maintained set when using them as analog inputs. I/O pins configured as analog input always read '0'.

INITIALIZING PORTD EXAMPLE 3-5:

BANKSEL	PORTD	;
CLRF	PORTD	;Init PORTD
BANKSEL	TRISD	;
MOVLW	B'00001100'	;Set RD<3:2> as inputs
MOVWF	TRISD	;and set RD<7:4,1:0>
		;as outputs

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3.6.1 RD<4:0>

Figure 3-19 shows the diagram for these pins. These pins are configured to function as general purpose I/O's.

Note:	RD<4:0> is available on PIC16F884/887
	only.

FIGURE 3-19: BLOCK DIAGRAM OF RD<4:0>



3.6.2 RD5/P1B⁽¹⁾

Figure 3-20 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- a PWM output

Note 1: RD5/P1B is available on PIC16F884/887 only. See RB2/AN8/P1B for this function on PIC16F882/883/886.

3.6.3 RD6/P1C⁽¹⁾

Figure 3-20 shows the diagram for this pin. This pin is configurable to function as one of the following:

• a general purpose I/O

a PWM output

Note 1:	RD6/P1C is available on PIC16F884/887
	only. See RB1/AN10/P1C/C12IN3- for
	this function on PIC16F882/883/886.

3.6.4 RD7/P1D⁽¹⁾

Figure 3-20 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- a PWM output

Note 1: RD7/P1D is available on PIC16F884/887 only. See RB4/AN11/P1D for this function on PIC16F882/883/886.



BLOCK DIAGRAM OF RD<7:5>



TABLE 3-4: SUMMARY OF REGISTERS ASSOCIATED WITH PORTD

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
PORTD	RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0	xxxx xxxx	uuuu uuuu
PSTRCON	_	_	—	STRSYNC	STRD	STRC	STRB	STRA	0 0001	0 0001
TRISD	TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0	1111 1111	1111 1111

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by PORTD.

3.7 **PORTE and TRISE Registers**

PORTE⁽¹⁾ is a 4-bit wide, bidirectional port. The corresponding data direction register is TRISE. Setting a TRISE bit (= 1) will make the corresponding PORTE pin an input (i.e., put the corresponding output driver in a High-Impedance mode). Clearing a TRISE bit (= 0) will make the corresponding PORTE pin an output (i.e., enable the output driver and put the contents of the output latch on the selected pin). The exception is RE3, which is input only and its TRIS bit will always read as '1'. Example 3-6 shows how to initialize PORTE.

Reading the PORTE register (Register 3-13) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch. RE3 reads '0' when MCLRE = 1.

Note 1:	RE<2:0>	pins	are	available	on
	PIC16F884	4/887 oi			

REGISTER 3-13: PORTE: PORTE REGISTER

The TRISE register (Register 3-14) controls the PORTE pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISE register are maintained set when using them as analog inputs. I/O pins configured as analog input always read '0'.

Note: The ANSEL register must be initialized to configure an analog channel as a digital input. Pins configured as analog inputs will read '0'.

EXAMPLE 3-6: **INITIALIZING PORTE**

BANKSEL	PORTE	;
CLRF	PORTE	;Init PORTE
BANKSEL	ANSEL	;
CLRF	ANSEL	;digital I/O
BCF	STATUS, RP1	;Bank 1
BANKSEL	TRISE	;
MOVLW	B`00001100'	;Set RE<3:2> as inputs
MOVWF	TRISE	;and set RE<1:0>
		;as outputs

U-0	U-0	U-0	U-0	R-x	R/W-x	R/W-x	R/W-x
—	—	—	—	RE3	RE2	RE1	RE0
bit 7	- -						bit 0

Legend:				
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown	

bit 3-0 RD<3:0>: PORTE General Purpose I/O Pin bit 1 = Port pin is > VIH0 = Port pin is < VIL

REGISTER 3-14: TRISE: PORTE TRI-STATE REGISTER

U-0	U-0	U-0	U-0	R-1 ⁽¹⁾	R/W-1	R/W-1	R/W-1
—	—	—	—	TRISE3	TRISE2	TRISE1	TRISE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-4 Unimplemented: Read as '0'

bit 3-0 TRISE<3:0>: PORTE Tri-State Control bit 1 = PORTE pin configured as an input (tri-stated) 0 = PORTE pin configured as an output

Note 1: TRISE<3> always reads '1'.

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3.7.1 RE0/AN5⁽¹⁾

This pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC

Note 1:	RE0/AN5 is available on PIC16F884/887
	only.

3.7.2 RE1/AN6⁽¹⁾

This pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC

Note 1: RE1/AN6 is available on PIC16F884/887 only.

3.7.3 RE2/AN7⁽¹⁾

This pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC

Note 1: RE2/AN7 is available on PIC16F884/887 only.

FIGURE 3-21: BLOCK DIAGRAM OF RE<2:0>



TABLE 3-5: SUMMARY OF REGISTERS ASSOCIATED WITH PORTE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ANSEL	ANS7	ANS6	ANS5	ANS4	ANS3	ANS2	ANS1	ANS0	1111 1111	1111 1111
PORTE		_		_	RE3	RE2	RE1	RE0	xxxx	uuuu
TRISE	_	_			TRISE3	TRISE2	TRISE1	TRISE0	1111	1111

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by PORTE

3.7.4 RE3/MCLR/VPP

Figure 3-22 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose input
- as Master Clear Reset with weak pull-up





4.0 OSCILLATOR MODULE (WITH FAIL-SAFE CLOCK MONITOR)

4.1 Overview

The Oscillator module has a wide variety of clock sources and selection features that allow it to be used in a wide range of applications while maximizing performance and minimizing power consumption. Figure 4-1 illustrates a block diagram of the Oscillator module.

Clock sources can be configured from external oscillators, quartz crystal resonators, ceramic resonators and Resistor-Capacitor (RC) circuits. In addition, the system clock source can be configured from one of two internal oscillators, with a choice of speeds selectable via software. Additional clock features include:

- Selectable system clock source between external or internal via software.
- Two-Speed Start-up mode, which minimizes latency between external oscillator start-up and code execution.
- Fail-Safe Clock Monitor (FSCM) designed to detect a failure of the external clock source (LP, XT, HS, EC or RC modes) and switch automatically to the internal oscillator.

The Oscillator module can be configured in one of eight clock modes.

- 1. EC External clock with I/O on OSC2/CLKOUT.
- 2. LP 32 kHz Low-Power Crystal mode.
- 3. XT Medium Gain Crystal or Ceramic Resonator Oscillator mode.
- 4. HS High Gain Crystal or Ceramic Resonator mode.
- 5. RC External Resistor-Capacitor (RC) with Fosc/4 output on OSC2/CLKOUT.
- 6. RCIO External Resistor-Capacitor (RC) with I/O on OSC2/CLKOUT.
- 7. INTOSC Internal oscillator with Fosc/4 output on OSC2 and I/O on OSC1/CLKIN.
- 8. INTOSCIO Internal oscillator with I/O on OSC1/CLKIN and OSC2/CLKOUT.

Clock Source modes are configured by the FOSC<2:0> bits in the Configuration Word Register 1 (CONFIG1). The internal clock can be generated from two internal oscillators. The HFINTOSC is a calibrated high-frequency oscillator. The LFINTOSC is an uncalibrated low-frequency oscillator.



FIGURE 4-1: PIC[®] MCU CLOCK SOURCE BLOCK DIAGRAM

4.2 Oscillator Control

The Oscillator Control (OSCCON) register (Figure 4-1) controls the system clock and frequency selection options. The OSCCON register contains the following bits:

- Frequency selection bits (IRCF)
- Frequency Status bits (HTS, LTS)
- System clock control bits (OSTS, SCS)

REGISTER 4-1: OSCCON: OSCILLATOR CONTROL REGISTER

U-0	R/W-1	R/W-1	R/W-0	R-1	R-0	R-0	R/W-0
—	IRCF2	IRCF1	IRCF0	OSTS ⁽¹⁾	HTS	LTS	SCS
bit 7	•						bit 0
Legend:							
R = Readable	bit	W = Writable	bit	U = Unimpler	mented bit, read	d as '0'	
-n = Value at F	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown
bit 7	Unimplemen	ted: Read as '	0'				
bit 6-4	IRCF<2:0>:	nternal Oscillat	or Frequency	Select bits			
	111 = 8 MHz						
	110 = 4 MHz	(default)					
	101 = 2 MHz 100 = 1 MHz						
	011 = 500 kH	łz					
	010 = 250 kH	lz					
	001 = 125 kH						
h it 0	000 = 31 kHz		rea aut Ctatura	L:₄(1)			
bit 3 OSIS: Oscillator Start-up Time-out Status bit''				register			
1 = Device is running from the internal oscillator (HFINTOSC or LFINTOSC)0 = Device is running from the internal oscillator (HFINTOSC or LFINTOSC)						register	
bit 2	HTS: HFINTO	OSC Status bit	(High Frequen	cy – 8 MHz to	125 kHz)		
	1 = HFINTOSC is stable						
	0 = HFINTOSC is not stable						
bit 1	LTS: LFINTOSC Stable bit (Low Frequency – 31 kHz)						
	1 = LFINTOS	SC is stable					
h # 0			;				
DITU	J = Internal of	I CIOCK Select I	Dit d for system of	lock			
	0 = Clock sol	urce defined by	/ FOSC<2:0> (of the CONFIG	1 register		
					3		

Note 1: Bit resets to '0' with Two-Speed Start-up and LP, XT or HS selected as the Oscillator mode or Fail-Safe mode is enabled.

4.3 Clock Source Modes

Clock Source modes can be classified as external or internal.

- External Clock modes rely on external circuitry for the clock source. Examples are: Oscillator modules (EC mode), quartz crystal resonators or ceramic resonators (LP, XT and HS modes) and Resistor-Capacitor (RC) mode circuits.
- Internal clock sources are contained internally within the Oscillator module. The Oscillator module has two internal oscillators: the 8 MHz High-Frequency Internal Oscillator (HFINTOSC) and the 31 kHz Low-Frequency Internal Oscillator (LFINTOSC).

The system clock can be selected between external or internal clock sources via the System Clock Select (SCS) bit of the OSCCON register. See **Section 4.6 "Clock Switching"** for additional information.

4.4 External Clock Modes

4.4.1 OSCILLATOR START-UP TIMER (OST)

If the Oscillator module is configured for LP, XT or HS modes, the Oscillator Start-up Timer (OST) counts 1024 oscillations from OSC1. This occurs following a Power-on Reset (POR) and when the Power-up Timer (PWRT) has expired (if configured), or a wake-up from Sleep. During this time, the program counter does not increment and program execution is suspended. The OST ensures that the oscillator circuit, using a quartz crystal resonator or ceramic resonator, has started and is providing a stable system clock to the Oscillator module. When switching between clock sources, a delay is required to allow the new clock to stabilize. These oscillator delays are shown in Table 4-1.

In order to minimize latency between external oscillator start-up and code execution, the Two-Speed Clock Start-up mode can be selected (see **Section 4.7 "Two-Speed Clock Start-up Mode"**).

Switch From	Switch To	Frequency	Oscillator Delay
Sleep/POR	LFINTOSC HFINTOSC	31 kHz 125 kHz to 8 MHz	Oscillator Warm-up Delay (TwaRM)
Sleep/POR	EC, RC	DC – 20 MHz	2 cycles
LFINTOSC (31 kHz)	EC, RC	DC – 20 MHz	1 cycle of each
Sleep/POR	LP, XT, HS	32 kHz to 20 MHz	1024 Clock Cycles (OST)
LFINTOSC (31 kHz)	HFINTOSC	125 kHz to 8 MHz	1 μs (approx.)

TABLE 4-1: OSCILLATOR DELAY EXAMPLES

4.4.2 EC MODE

The External Clock (EC) mode allows an externally generated logic level as the system clock source. When operating in this mode, an external clock source is connected to the OSC1 input and the OSC2 is available for general purpose I/O. Figure 4-2 shows the pin connections for EC mode.

The Oscillator Start-up Timer (OST) is disabled when EC mode is selected. Therefore, there is no delay in operation after a Power-on Reset (POR) or wake-up from Sleep. Because the PIC[®] MCU design is fully static, stopping the external clock input will have the effect of halting the device while leaving all data intact. Upon restarting the external clock, the device will resume operation as if no time had elapsed.

FIGURE 4-2:

EXTERNAL CLOCK (EC) MODE OPERATION



4.4.3 LP, XT, HS MODES

The LP, XT and HS modes support the use of quartz crystal resonators or ceramic resonators connected to OSC1 and OSC2 (Figure 4-3). The mode selects a low, medium or high gain setting of the internal inverter-amplifier to support various resonator types and speed.

LP Oscillator mode selects the lowest gain setting of the internal inverter-amplifier. LP mode current consumption is the least of the three modes. This mode is designed to drive only 32.768 kHz tuning-fork type crystals (watch crystals).

XT Oscillator mode selects the intermediate gain setting of the internal inverter-amplifier. XT mode current consumption is the medium of the three modes. This mode is best suited to drive resonators with a medium drive level specification.

HS Oscillator mode selects the highest gain setting of the internal inverter-amplifier. HS mode current consumption is the highest of the three modes. This mode is best suited for resonators that require a high drive setting.

Figure 4-3 and Figure 4-4 show typical circuits for quartz crystal and ceramic resonators, respectively.





2: The value of RF varies with the Oscillator mode selected (typically between 2 M Ω to 10 M Ω).

- **Note 1:** Quartz crystal characteristics vary according to type, package and manufacturer. The user should consult the manufacturer data sheets for specifications and recommended application.
 - 2: Always verify oscillator performance over the VDD and temperature range that is expected for the application.
 - **3:** For oscillator design assistance, reference the following Microchip Applications Notes:
 - AN826, "Crystal Oscillator Basics and Crystal Selection for rfPIC[®] and PIC[®] Devices" (DS00826)
 - AN849, "Basic PIC[®] Oscillator Design" (DS00849)
 - AN943, "Practical PIC[®] Oscillator Analysis and Design" (DS00943)
 - AN949, "Making Your Oscillator Work" (DS00949)





operation.

4.4.4 EXTERNAL RC MODES

The external Resistor-Capacitor (RC) modes support the use of an external RC circuit. This allows the designer maximum flexibility in frequency choice while keeping costs to a minimum when clock accuracy is not required. There are two modes: RC and RCIO.

In RC mode, the RC circuit connects to OSC1. OSC2/CLKOUT outputs the RC oscillator frequency divided by 4. This signal may be used to provide a clock for external circuitry, synchronization, calibration, test or other application requirements. Figure 4-5 shows the external RC mode connections.

Vdd PIC[®] MCU REXT OSC1/CLKIN Internal Clock CEXT Vss -OSC2/CLKOUT(1) Fosc/4 or I/O⁽²⁾ Recommended values: $10 \text{ k}\Omega \leq \text{REXT} \leq 100 \text{ k}\Omega$. <3V $3 \text{ k}\Omega \leq \text{Rext} \leq 100 \text{ k}\Omega, 3-5 \text{V}$ CEXT > 20 pF, 2-5V Alternate pin functions are listed in the Note 1: Section 1.0 "Device Overview". 2: Output depends upon RC or RCIO Clock mode.

FIGURE 4-5: EXTERNAL RC MODES

In RCIO mode, the RC circuit is connected to OSC1. OSC2 becomes an additional general purpose I/O pin.

The RC oscillator frequency is a function of the supply voltage, the resistor (REXT) and capacitor (CEXT) values and the operating temperature. Other factors affecting the oscillator frequency are:

- threshold voltage variation
- component tolerances
- packaging variations in capacitance

The user also needs to take into account variation due to tolerance of external RC components used.

4.5 Internal Clock Modes

The Oscillator module has two independent, internal oscillators that can be configured or selected as the system clock source.

- 1. The **HFINTOSC** (High-Frequency Internal Oscillator) is factory calibrated and operates at 8 MHz. The frequency of the HFINTOSC can be user-adjusted via software using the OSCTUNE register (Register 4-2).
- 2. The **LFINTOSC** (Low-Frequency Internal Oscillator) is uncalibrated and operates at 31 kHz.

The system clock speed can be selected via software using the Internal Oscillator Frequency Select bits IRCF<2:0> of the OSCCON register.

The system clock can be selected between external or internal clock sources via the System Clock Selection (SCS) bit of the OSCCON register. See **Section 4.6** "**Clock Switching**" for more information.

4.5.1 INTOSC AND INTOSCIO MODES

The INTOSC and INTOSCIO modes configure the internal oscillators as the system clock source when the device is programmed using the oscillator selection or the FOSC<2:0> bits in the Configuration Word Register 1 (CONFIG1).

In **INTOSC** mode, OSC1/CLKIN is available for general purpose I/O. OSC2/CLKOUT outputs the selected internal oscillator frequency divided by 4. The CLKOUT signal may be used to provide a clock for external circuitry, synchronization, calibration, test or other application requirements.

In **INTOSCIO** mode, OSC1/CLKIN and OSC2/CLKOUT are available for general purpose I/O.

4.5.2 HFINTOSC

The High-Frequency Internal Oscillator (HFINTOSC) is a factory calibrated 8 MHz internal clock source. The frequency of the HFINTOSC can be altered via software using the OSCTUNE register (Register 4-2).

The output of the HFINTOSC connects to a postscaler and multiplexer (see Figure 4-1). One of seven frequencies can be selected via software using the IRCF<2:0> bits of the OSCCON register. See **Section 4.5.4 "Frequency Select Bits (IRCF)**" for more information.

The HFINTOSC is enabled by selecting any frequency between 8 MHz and 125 kHz by setting the IRCF<2:0> bits of the OSCCON register \neq 000. Then, set the System Clock Source (SCS) bit of the OSCCON register to '1' or enable Two-Speed Start-up by setting the IESO bit in the Configuration Word Register 1 (CONFIG1) to '1'.

The HF Internal Oscillator (HTS) bit of the OSCCON register indicates whether the HFINTOSC is stable or not.

4.5.2.1 OSCTUNE Register

The HFINTOSC is factory calibrated but can be adjusted in software by writing to the OSCTUNE register (Register 4-2).

The default value of the OSCTUNE register is '0'. The value is a 5-bit two's complement number.

When the OSCTUNE register is modified, the HFINTOSC frequency will begin shifting to the new frequency. Code execution continues during this shift. There is no indication that the shift has occurred.

OSCTUNE does not affect the LFINTOSC frequency. Operation of features that depend on the LFINTOSC clock source frequency, such as the Power-up Timer (PWRT), Watchdog Timer (WDT), Fail-Safe Clock Monitor (FSCM) and peripherals, are *not* affected by the change in frequency.

REGISTER 4-2: OSCTUNE: OSCILLATOR TUNING REGISTER

U-0	U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
_	—	—	TUN4	TUN3	TUN2	TUN1	TUN0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-5	Unimplemented: Read as '0'
bit 4-0	TUN<4:0>: Frequency Tuning bits
	01111 = Maximum frequency
	01110 =
	•
	•
	•
	00001 = 00000 = Oscillator module is running at the calibrated frequency. 11111 =
	•
	•
	•
	10000 = Minimum frequency

4.5.3 LFINTOSC

The Low-Frequency Internal Oscillator (LFINTOSC) is an uncalibrated 31 kHz internal clock source.

The output of the LFINTOSC connects to a postscaler and multiplexer (see Figure 4-1). Select 31 kHz, via software, using the IRCF<2:0> bits of the OSCCON register. See **Section 4.5.4** "**Frequency Select Bits (IRCF)**" for more information. The LFINTOSC is also the frequency for the Power-up Timer (PWRT), Watchdog Timer (WDT) and Fail-Safe Clock Monitor (FSCM).

The LFINTOSC is enabled by selecting 31 kHz (IRCF<2:0> bits of the OSCCON register = 000) as the system clock source (SCS bit of the OSCCON register = 1), or when any of the following are enabled:

- Two-Speed Start-up IESO bit of the Configuration Word Register 1 = 1 and IRCF<2:0> bits of the OSCCON register = 000
- Power-up Timer (PWRT)
- Watchdog Timer (WDT)
- Fail-Safe Clock Monitor (FSCM)

The LF Internal Oscillator (LTS) bit of the OSCCON register indicates whether the LFINTOSC is stable or not.

4.5.4 FREQUENCY SELECT BITS (IRCF)

The output of the 8 MHz HFINTOSC and 31 kHz LFINTOSC connects to a postscaler and multiplexer (see Figure 4-1). The Internal Oscillator Frequency Select bits IRCF<2:0> of the OSCCON register select the frequency output of the internal oscillators. One of eight frequencies can be selected via software:

- 8 MHz
- 4 MHz (Default after Reset)
- 2 MHz
- 1 MHz
- 500 kHz
- 250 kHz
- 125 kHz
- 31 kHz (LFINTOSC)

Note:	Following any Reset, the IRCF<2:0> bits of
	the OSCCON register are set to '110' and
	the frequency selection is set to 4 MHz.
	The user can modify the IRCF bits to
	select a different frequency.

4.5.5 HFINTOSC AND LFINTOSC CLOCK SWITCH TIMING

When switching between the LFINTOSC and the HFINTOSC, the new oscillator may already be shut down to save power (see Figure 4-6). If this is the case, there is a delay after the IRCF<2:0> bits of the OSCCON register are modified before the frequency selection takes place. The LTS and HTS bits of the OSCCON register will reflect the current active status of the LFINTOSC and HFINTOSC oscillators. The timing of a frequency selection is as follows:

- 1. IRCF<2:0> bits of the OSCCON register are modified.
- 2. If the new clock is shut down, a clock start-up delay is started.
- 3. Clock switch circuitry waits for a falling edge of the current clock.
- 4. CLKOUT is held low and the clock switch circuitry waits for a rising edge in the new clock.
- CLKOUT is now connected with the new clock. LTS and HTS bits of the OSCCON register are updated as required.
- 6. Clock switch is complete.

See Figure 4-1 for more details.

If the internal oscillator speed selected is between 8 MHz and 125 kHz, there is no start-up delay before the new frequency is selected. This is because the old and new frequencies are derived from the HFINTOSC via the postscaler and multiplexer.

Start-up delay specifications are located in the oscillator tables of **Section 17.0** "**Electrical Specifications**".

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FIGURE 4-6:	INTERNAL OSCILLATOR SWITCH TIMING
HFINTOSC -+ L	FINTOSC (FSCM and WDT disabled)
HFINTOSC	Start-up Time 2-cycle Sync Running
LFINTOSC	
IRCF <2:0>	$\neq 0$ $X = 0$
System Clock	
HFINTOSC → L	FINTOSC (Either FSCM or WDT enabled)
HFINTOSC	2-cycle Sync Running
LFINTOSC -	
IRCF <2:0>	$\neq 0$ $\lambda = 0$
System Clock	
LFINTOSC -> F	IFINTOSC
LFINTOSC	EFINIOSC turns off unless WDT or FSCM is enabled
HFINTOSC	
IRCF <2:0>	= 0 1/4 0
System Clock	

4.6 Clock Switching

The system clock source can be switched between external and internal clock sources via software using the System Clock Select (SCS) bit of the OSCCON register.

4.6.1 SYSTEM CLOCK SELECT (SCS) BIT

The System Clock Select (SCS) bit of the OSCCON register selects the system clock source that is used for the CPU and peripherals.

- When the SCS bit of the OSCCON register = 0, the system clock source is determined by configuration of the FOSC<2:0> bits in the Configuration Word Register 1 (CONFIG1).
- When the SCS bit of the OSCCON register = 1, the system clock source is chosen by the internal oscillator frequency selected by the IRCF<2:0> bits of the OSCCON register. After a Reset, the SCS bit of the OSCCON register is always cleared.
- Note: Any automatic clock switch, which may occur from Two-Speed Start-up or Fail-Safe Clock Monitor, does not update the SCS bit of the OSCCON register. The user can monitor the OSTS bit of the OSCCON register to determine the current system clock source.

4.6.2 OSCILLATOR START-UP TIME-OUT STATUS (OSTS) BIT

The Oscillator Start-up Time-out Status (OSTS) bit of the OSCCON register indicates whether the system clock is running from the external clock source, as defined by the FOSC<2:0> bits in the Configuration Word Register 1 (CONFIG1), or from the internal clock source. In particular, OSTS indicates that the Oscillator Start-up Timer (OST) has timed out for LP, XT or HS modes.

4.7 Two-Speed Clock Start-up Mode

Two-Speed Start-up mode provides additional power savings by minimizing the latency between external oscillator start-up and code execution. In applications that make heavy use of the Sleep mode, Two-Speed Start-up will remove the external oscillator start-up time from the time spent awake and can reduce the overall power consumption of the device.

This mode allows the application to wake-up from Sleep, perform a few instructions using the INTOSC as the clock source and go back to Sleep without waiting for the primary oscillator to become stable.

Note: Executing a SLEEP instruction will abort the oscillator start-up time and will cause the OSTS bit of the OSCCON register to remain clear. When the Oscillator module is configured for LP, XT or HS modes, the Oscillator Start-up Timer (OST) is enabled (see **Section 4.4.1 "Oscillator Start-up Timer (OST)"**). The OST will suspend program execution until 1024 oscillations are counted. Two-Speed Start-up mode minimizes the delay in code execution by operating from the internal oscillator as the OST is counting. When the OST count reaches 1024 and the OSTS bit of the OSCCON register is set, program execution switches to the external oscillator.

4.7.1 TWO-SPEED START-UP MODE CONFIGURATION

Two-Speed Start-up mode is configured by the following settings:

- IESO (of the Configuration Word Register 1) = 1; Internal/External Switchover bit (Two-Speed Start-up mode enabled).
- SCS (of the OSCCON register) = 0.
- FOSC<2:0> bits in the Configuration Word Register 1 (CONFIG1) configured for LP, XT or HS mode.

Two-Speed Start-up mode is entered after:

- Power-on Reset (POR) and, if enabled, after Power-up Timer (PWRT) has expired, or
- Wake-up from Sleep.

If the external clock oscillator is configured to be anything other than LP, XT or HS mode, then Two-speed Start-up is disabled. This is because the external clock oscillator does not require any stabilization time after POR or an exit from Sleep.

4.7.2 TWO-SPEED START-UP SEQUENCE

- 1. Wake-up from Power-on Reset or Sleep.
- Instructions begin execution by the internal oscillator at the frequency set in the IRCF<2:0> bits of the OSCCON register.
- 3. OST enabled to count 1024 clock cycles.
- 4. OST timed out, wait for falling edge of the internal oscillator.
- 5. OSTS is set.
- 6. System clock held low until the next falling edge of new clock (LP, XT or HS mode).
- 7. System clock is switched to external clock source.

4.7.3 CHECKING TWO-SPEED CLOCK STATUS

Checking the state of the OSTS bit of the OSCCON register will confirm if the microcontroller is running from the external clock source, as defined by the FOSC<2:0> bits in the Configuration Word Register 1 (CONFIG1), or the internal oscillator.

FIGURE 4-7:	TWO-SPEED START-UP	
HFINTOSC		
OSC1		
OSC2		
Program Counter	PC - N { PC + 1	-

1

System Clock
4.8 Fail-Safe Clock Monitor

The Fail-Safe Clock Monitor (FSCM) allows the device to continue operating should the external oscillator fail. The FSCM can detect oscillator failure any time after the Oscillator Start-up Timer (OST) has expired. The FSCM is enabled by setting the FCMEN bit in the Configuration Word Register 1 (CONFIG1). The FSCM is applicable to all external Oscillator modes (LP, XT, HS, EC, RC and RCIO).

FIGURE 4-8: FSCM BLOCK DIAGRAM



4.8.1 FAIL-SAFE DETECTION

The FSCM module detects a failed oscillator by comparing the external oscillator to the FSCM sample clock. The sample clock is generated by dividing the LFINTOSC by 64. See Figure 4-8. Inside the fail detector block is a latch. The external clock sets the latch on each falling edge of the external clock. The sample clock clears the latch on each rising edge of the sample clock. A failure is detected when an entire half-cycle of the sample clock elapses before the primary clock goes low.

4.8.2 FAIL-SAFE OPERATION

When the external clock fails, the FSCM switches the device clock to an internal clock source and sets the bit flag OSFIF of the PIR2 register. Setting this flag will generate an interrupt if the OSFIE bit of the PIE2 register is also set. The device firmware can then take steps to mitigate the problems that may arise from a failed clock. The system clock will continue to be sourced from the internal clock source until the device firmware successfully restarts the external oscillator and switches back to external operation.

The internal clock source chosen by the FSCM is determined by the IRCF<2:0> bits of the OSCCON register. This allows the internal oscillator to be configured before a failure occurs.

4.8.3 FAIL-SAFE CONDITION CLEARING

The Fail-Safe condition is cleared after a Reset, executing a SLEEP instruction or toggling the SCS bit of the OSCCON register. When the SCS bit is toggled, the OST is restarted. While the OST is running, the device continues to operate from the INTOSC selected in OSCCON. When the OST times out, the Fail-Safe condition is cleared and the device will be operating from the external clock source. The Fail-Safe condition must be cleared before the OSFIF flag can be cleared.

4.8.4 RESET OR WAKE-UP FROM SLEEP

The FSCM is designed to detect an oscillator failure after the Oscillator Start-up Timer (OST) has expired. The OST is used after waking up from Sleep and after any type of Reset. The OST is not used with the EC or RC Clock modes so that the FSCM will be active as soon as the Reset or wake-up has completed. When the FSCM is enabled, the Two-Speed Start-up is also enabled. Therefore, the device will always be executing code while the OST is operating.

Note:	Due to the wide range of oscillator start-up times, the Fail-Safe circuit is not active during oscillator start-up (i.e., after exiting Reset or Sleep). After an appropriate amount of time, the user should check the OSTS bit of the OSCCON register to verify the oscillator start-up and that the system clock switchover has successfully
	completed.

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FIGURE 4-9: FSCM TIMING DIAGRAM



Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets ⁽¹⁾
CONFIG1(2)	CPD	CP	MCLRE	PWRTE	WDTE	FOSC2	FOSC1	FOSC0		
OSCCON	—	IRCF2	IRCF1	IRCF0	OSTS	HTS	LTS	SCS	-110 x000	-110 x000
OSCTUNE	_	—	—	TUN4	TUN3	TUN2	TUN1	TUN0	0 0000	u uuuu
PIE2	OSFIE	C2IE	C1IE	EEIE	BCLIE	ULPWUIE	—	CCP2IE	0000 00-0	0000 00-0
PIR2	OSFIF	C2IF	C1IF	EEIF	BCLIF	ULPWUIF	_	CCP2IF	0000 00-0	0000 00-0

Note 1: Other (non Power-up) Resets include MCLR Reset and Watchdog Timer Reset during normal operation.

2: See Configuration Word Register 1 (Register 14-1) for operation of all register bits.

5.0 TIMER0 MODULE

The Timer0 module is an 8-bit timer/counter with the following features:

- 8-bit timer/counter register (TMR0)
- 8-bit prescaler (shared with Watchdog Timer)
- Programmable internal or external clock source
- · Programmable external clock edge selection
- Interrupt on overflow

Figure 5-1 is a block diagram of the Timer0 module.

5.1 Timer0 Operation

When used as a timer, the Timer0 module can be used as either an 8-bit timer or an 8-bit counter.

5.1.1 8-BIT TIMER MODE

When used as a timer, the Timer0 module will increment every instruction cycle (without prescaler). Timer mode is selected by clearing the T0CS bit of the OPTION register to '0'.

When TMR0 is written, the increment is inhibited for two instruction cycles immediately following the write.

Note: The value written to the TMR0 register can be adjusted, in order to account for the two instruction cycle delay when TMR0 is written.

5.1.2 8-BIT COUNTER MODE

When used as a counter, the Timer0 module will increment on every rising or falling edge of the T0CKI pin. The incrementing edge is determined by the T0SE bit of the OPTION register. Counter mode is selected by setting the T0CS bit of the OPTION register to '1'.



FIGURE 5-1: TIMER0/WDT PRESCALER BLOCK DIAGRAM

5.1.3 SOFTWARE PROGRAMMABLE PRESCALER

A single software programmable prescaler is available for use with either Timer0 or the Watchdog Timer (WDT), but not both simultaneously. The prescaler assignment is controlled by the PSA bit of the OPTION register. To assign the prescaler to Timer0, the PSA bit must be cleared to a '0'.

There are 8 prescaler options for the Timer0 module ranging from 1:2 to 1:256. The prescale values are selectable via the PS<2:0> bits of the OPTION register. In order to have a 1:1 prescaler value for the Timer0 module, the prescaler must be assigned to the WDT module.

The prescaler is not readable or writable. When assigned to the Timer0 module, all instructions writing to the TMR0 register will clear the prescaler.

When the prescaler is assigned to WDT, a CLRWDT instruction will clear the prescaler along with the WDT.

5.1.3.1 Switching Prescaler Between Timer0 and WDT Modules

As a result of having the prescaler assigned to either Timer0 or the WDT, it is possible to generate an unintended device Reset when switching prescaler values. When changing the prescaler assignment from Timer0 to the WDT module, the instruction sequence shown in Example 5-1, must be executed.

EXAMPLE 5-1: CHANGING PRESCALER (TIMER0 \rightarrow WDT)

BANKSEL CLRWDT	TMR0	; ;Clear WDT
CLRF	TMR0	;Clear TMR0 and ;prescaler
BANKSEL	OPTION_REG	;
BSF	OPTION_REG, PSA	;Select WDT
CLRWDT		i
		;
MOVLW	b'11111000'	;Mask prescaler
ANDWF	OPTION_REG,W	;bits
IORLW	b'00000101'	;Set WDT prescaler
MOVWF	OPTION_REG	;to 1:32

When changing the prescaler assignment from the WDT to the Timer0 module, the following instruction sequence must be executed (see Example 5-2).

EXAMPLE 5-2:	CHANGING PRESCALER
	(WDT \rightarrow TIMER0)

CLRWDT		;Clear WDT and ;prescaler
BANKSEL	OPTION REG	;
MOVLW	b'11110000'	;Mask TMR0 select and
ANDWF	OPTION_REG,W	;prescaler bits
IORLW	b'0000011'	;Set prescale to 1:16
MOVWF	OPTION_REG	;

5.1.4 TIMER0 INTERRUPT

Timer0 will generate an interrupt when the TMR0 register overflows from FFh to 00h. The T0IF interrupt flag bit of the INTCON register is set every time the TMR0 register overflows, regardless of whether or not the Timer0 interrupt is enabled. The T0IF bit must be cleared in software. The Timer0 interrupt enable is the T0IE bit of the INTCON register.

Note:	The Timer0 interrupt cannot wake the
	processor from Sleep since the timer is frozen during Sleep.
	5 .

5.1.5 USING TIMER0 WITH AN EXTERNAL CLOCK

When Timer0 is in Counter mode, the synchronization of the T0CKI input and the Timer0 register is accomplished by sampling the prescaler output on the Q2 and Q4 cycles of the internal phase clocks. Therefore, the high and low periods of the external clock source must meet the timing requirements as shown in the **Section 17.0 "Electrical Specifications"**.

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
RBPU	INTEDG	TOCS	TOSE	PSA	PS2	PS1	PS0
bit 7	·						bit 0
Legend:							
R = Readable	e bit	W = Writable	bit	U = Unimplei	mented bit, rea	ad as '0'	
-n = Value at I	POR	'1' = Bit is se	t	'0' = Bit is cle	eared	x = Bit is unki	nown
bit 7	RBPU: POR	TB Pull-up Ena	able bit				
	1 = PORTB	pull-ups are dis	abled				
	0 = PORTB p	pull-ups are en	abled by indivi	idual PORT late	ch values		
bit 6	INTEDG: Inte	errupt Edge Se	lect bit				
	1 = Interrupt	on rising edge	of INT pin				
h:4 C		On failing edge					
DIES		D Clock Source	Select bit				
	1 = Internal i	nstruction cycle	e clock (Fosc/	4)			
bit 4	TOSE: TMRC) Source Edge	Select bit	•)			
	1 = Incremer	nt on high-to-lo	w transition or	n TOCKI pin			
	0 = Incremer	nt on low-to-hig	h transition or	TOCKI pin			
bit 3	PSA: Presca	aler Assignmen	t bit				
	1 = Prescale	r is assigned to	the WDT				
	0 = Prescale	r is assigned to	the Timer0 m	nodule			
bit 2-0	PS<2:0>: Pr	escaler Rate S	elect bits				
	BIT	VALUE TMR0 F	RATE WDT RA	TE			
		000 1:2	1:1				
	(001 1:4	1:2				
		010 1:8	1:4				
			0 1:8 2 1.46				
		100 1:3	∠ 1:16 4 1:32				
		110 1 · 1	28 1.64				

REGISTER 5-1: OPTION_REG: OPTION REGISTER

Note 1: A dedicated 16-bit WDT postscaler is available. See Section 14.5 "Watchdog Timer (WDT)" for more information.

1 : 128

TABLE 5-1: SUMMARY OF REGISTERS ASSOCIATED WITH TIMER0

1 : 256

111

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR		Value on POR, BOR Res		e on other sets
TMR0	Timer0 N	limer0 Module Register								xxxx	uuuu	uuuu	
INTCON	GIE	PEIE	TOIE	INTE	RBIE	T0IF	INTF	RBIF	0000	000x	0000	000x	
OPTION_REG	RBPU	INTEDG	TOCS	T0SE	PSA	PS2	PS1	PS0	1111	1111	1111	1111	
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	1111	1111	1111	1111	

Legend: – = Unimplemented locations, read as '0', u = unchanged, x = unknown. Shaded cells are not used by the Timer0 module.

6.0 TIMER1 MODULE WITH GATE CONTROL

The Timer1 module is a 16-bit timer/counter with the following features:

- 16-bit timer/counter register pair (TMR1H:TMR1L)
- · Programmable internal or external clock source
- 3-bit prescaler
- Optional LP oscillator
- · Synchronous or asynchronous operation
- Timer1 gate (count enable) via comparator or T1G pin
- Interrupt on overflow
- Wake-up on overflow (external clock, Asynchronous mode only)
- Time base for the Capture/Compare function
- Special Event Trigger (with ECCP)
- Comparator output synchronization to Timer1 clock

Figure 6-1 is a block diagram of the Timer1 module.

TMR1GE T1GINV TMR10N Set flag bit TMR1IF on To C2 Comparator Module Timer1 Clock Overflow TMR1⁽²⁾ Synchronized E١ clock input TMR1H TMR1L T1 OSC T1SYNC T10SI 1 Synchronize⁽³⁾ Prescaler 0 1, 2, 4, 8 🖊 det Fosc/4 T10S0 📉 Internal 0 / 2 Clock Sleep input T1CKPS<1:0> T1OSCEN TMR1CS T1G 1 C2OUT 0 т1СКІ T1GSS Note 1: ST Buffer is low power type when using LP oscillator, or high speed type when using T1CKI. 2: Timer1 register increments on rising edge. Synchronize does not operate while in Sleep. 3:

FIGURE 6-1: TIMER1 BLOCK DIAGRAM

6.1 Timer1 Operation

The Timer1 module is a 16-bit incrementing counter which is accessed through the TMR1H:TMR1L register pair. Writes to TMR1H or TMR1L directly update the counter.

When used with an internal clock source, the module is a timer. When used with an external clock source, the module can be used as either a timer or counter.

6.2 Clock Source Selection

The TMR1CS bit of the T1CON register is used to select the clock source. When TMR1CS = 0, the clock source is FOSC/4. When TMR1CS = 1, the clock source is supplied externally.

Clock Source	TMR1CS
Fosc/4	0
T1CKI pin	1

6.2.1 INTERNAL CLOCK SOURCE

When the internal clock source is selected the TMR1H:TMR1L register pair will increment on multiples of Fosc as determined by the Timer1 prescaler.

6.2.2 EXTERNAL CLOCK SOURCE

When the external clock source is selected, the Timer1 module may work as a timer or a counter.

When counting, Timer1 is incremented on the rising edge of the external clock input T1CKI. In addition, the Counter mode clock can be synchronized to the microcontroller system clock or run asynchronously.

If an external clock oscillator is needed (and the microcontroller is using the INTOSC without CLKOUT), Timer1 can use the LP oscillator as a clock source.

Note:	In Counter mode, a falling edge must be
	registered by the counter prior to the first
	incrementing rising edge.

6.3 Timer1 Prescaler

Timer1 has four prescaler options allowing 1, 2, 4 or 8 divisions of the clock input. The T1CKPS bits of the T1CON register control the prescale counter. The prescale counter is not directly readable or writable; however, the prescaler counter is cleared upon a write to TMR1H or TMR1L.

6.4 Timer1 Oscillator

A low-power 32.768 kHz crystal oscillator is built-in between pins T1OSI (input) and T1OSO (amplifier output). The oscillator is enabled by setting the T1OSCEN control bit of the T1CON register. The oscillator will continue to run during Sleep.

The Timer1 oscillator is identical to the LP oscillator. The user must provide a software time delay to ensure proper oscillator start-up.

TRISC0 and TRISC1 bits are set when the Timer1 oscillator is enabled. RC0 and RC1 bits read as '0' and TRISC0 and TRISC1 bits read as '1'.

Note:	The oscillator requires a start-up and
	stabilization time before use. Thus,
	T1OSCEN should be set and a suitable
	delay observed prior to enabling Timer1.

6.5 Timer1 Operation in Asynchronous Counter Mode

If control bit T1SYNC of the T1CON register is set, the external clock input is not synchronized. The timer continues to increment asynchronous to the internal phase clocks. The timer will continue to run during Sleep and can generate an interrupt on overflow, which will wake-up the processor. However, special precautions in software are needed to read/write the timer (see Section 6.5.1 "Reading and Writing Timer1 in Asynchronous Counter Mode").

Note: When switching from synchronous to asynchronous operation, it is possible to skip an increment. When switching from asynchronous to synchronous operation, it is possible to produce a single spurious increment.

6.5.1 READING AND WRITING TIMER1 IN ASYNCHRONOUS COUNTER MODE

Reading TMR1H or TMR1L while the timer is running from an external asynchronous clock will ensure a valid read (taken care of in hardware). However, the user should keep in mind that reading the 16-bit timer in two 8-bit values itself, poses certain problems, since the timer may overflow between the reads.

For writes, it is recommended that the user simply stop the timer and write the desired values. A write contention may occur by writing to the timer registers, while the register is incrementing. This may produce an unpredictable value in the TMR1H:TTMR1L register pair.

6.6 Timer1 Gate

Timer1 gate source is software configurable to be the T1G pin or the output of Comparator C2. This allows the device to directly time external events using T1G or analog events using Comparator C2. See the CM2CON1 register (Register 8-3) for selecting the Timer1 gate source. This feature can simplify the software for a Delta-Sigma A/D converter and many other applications. For more information on Delta-Sigma A/D converters, see the Microchip web site (www.microchip.com).

Note:	TMR1GE bit of the T1CON register must
	be set to use either T1G or C2OUT as the
	Timer1 gate source. See Register 8-3 for
	more information on selecting the Timer1
	gate source.

Timer1 gate can be inverted using the T1GINV bit of the T1CON register, whether it originates from the T1G pin or Comparator C2 output. This configures Timer1 to measure either the active-high or active-low time between events.

6.7 Timer1 Interrupt

The Timer1 register pair (TMR1H:TMR1L) increments to FFFFh and rolls over to 0000h. When Timer1 rolls over, the Timer1 interrupt flag bit of the PIR1 register is set. To enable the interrupt on rollover, you must set these bits:

- Timer1 interrupt enable bit of the PIE1 register
- PEIE bit of the INTCON register
- GIE bit of the INTCON register

The interrupt is cleared by clearing the TMR1IF bit in the Interrupt Service Routine.

Note: The TMR1H:TTMR1L register pair and the TMR1IF bit should be cleared before enabling interrupts.

6.8 Timer1 Operation During Sleep

Timer1 can only operate during Sleep when setup in Asynchronous Counter mode. In this mode, an external crystal or clock source can be used to increment the counter. To set up the timer to wake the device:

- TMR1ON bit of the T1CON register must be set
- TMR1IE bit of the PIE1 register must be set
- PEIE bit of the INTCON register must be set

The device will wake-up on an overflow and execute the next instruction. If the GIE bit of the INTCON register is set, the device will call the Interrupt Service Routine (0004h).

6.9 ECCP Capture/Compare Time Base

The ECCP module uses the TMR1H:TMR1L register pair as the time base when operating in Capture or Compare mode.

In Capture mode, the value in the TMR1H:TMR1L register pair is copied into the CCPRxH:CCPRxL register pair on a configured event.

In Compare mode, an event is triggered when the value CCPRxH:CCPRxL register pair matches the value in the TMR1H:TMR1L register pair. This event can be a Special Event Trigger.

See Section 11.0 "Capture/Compare/PWM Modules (CCP1 and CCP2)" for more information.

6.10 ECCP Special Event Trigger

If an ECCP is configured to trigger a special event, the trigger will clear the TMR1H:TMR1L register pair. This special event does not cause a Timer1 interrupt. The ECCP module may still be configured to generate a ECCP interrupt.

In this mode of operation, the CCPRxH:CCPRxL register pair effectively becomes the period register for Timer1.

Timer1 should be synchronized to the FOSC to utilize the Special Event Trigger. Asynchronous operation of Timer1 can cause a Special Event Trigger to be missed.

In the event that a write to TMR1H or TMR1L coincides with a Special Event Trigger from the ECCP, the write will take precedence.

For more information, see Section 11.0 "Capture/Compare/PWM Modules (CCP1 and CCP2)".

6.11 Comparator Synchronization

The same clock used to increment Timer1 can also be used to synchronize the comparator output. This feature is enabled in the Comparator module.

When using the comparator for Timer1 gate, the comparator output should be synchronized to Timer1. This ensures Timer1 does not miss an increment if the comparator changes.

For more information, see **Section 8.0 "Comparator Module"**.



FIGURE 6-2: TIMER1 INCREMENTING EDGE

6.12 Timer1 Control Register

The Timer1 Control register (T1CON), shown in Register 6-1, is used to control Timer1 and select the various features of the Timer1 module.

REGISTER 6-1: T1CON: TIMER1 CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
T1GINV	⁽¹⁾ TMR1GE ⁽²⁾	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N
bit 7							bit 0
Legend:							
R = Reada	able bit	W = Writable	bit	U = Unimplen	nented bit, read	d as '0'	
-n = Value	at POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown
bit 7	T1GINV: Time	er1 Gate Invert	bit (1)				
	1 = Timer1 ga	ite is active-hig ite is active-lov	n (Timer1 cou v (Timer1 cour	nts when gate i	is nign) s low)		
bit 6	TMR1GF· Tim	ner1 Gate Ena	ble bit(2)	to mion gate to			
5100	<u>If TMR10N =</u>	<u>0:</u>					
	This bit is igno	ored					
	<u>If TMR10N =</u> 1 – Timer1 is	<u>1:</u> on if Timer1 a:	ate is not active	2			
	0 = Timer1 is	on		5			
bit 5-4	T1CKPS<1:0	>: Timer1 Inpu	t Clock Presca	le Select bits			
	11 = 1:8 Pres	cale Value					
	10 = 1:4 Pres	cale Value					
	00 = 1:1 Pres	cale Value					
bit 3	T1OSCEN: LI	P Oscillator En	able Control b	it			
	1 = LP oscilla	tor is enabled	for Timer1 cloc	:k			
1.11.0	0 = LP oscillat	tor is off			6 11 2		
bit 2	TISYNC: Tim	er1 External C	lock Input Syn	chronization C	ontrol bit		
	$\frac{1101 \times 105 = 1}{1 = \text{Do not svi}}$	nchronize exte	rnal clock inpu	ıt			
	0 = Synchroni	ize external clo	ock input				
	$\frac{\text{TMR1CS} = 0}{This bit is ison$	arad Timort	non the interne	l olook			
hit 1		ort Clock Sou	ses the interna	II CIUCK			
	1 = External o	lock from T1C	KI pin (on the	risina edae)			
	0 = Internal cl	ock (Fosc/4)		g odgoj			
bit 0	TMR1ON: Tin	ner1 On bit					
	1 = Enables T	imer1					
	0 = Stops Tim	er1					
Note 1:	T1GINV bit inverts	the Timer1 ga	te logic, regard	dless of source		T4000 1 16 4 1	0100011
2:	IMR1GE bit must	be set to use e	either 11G bin (or C2OUT. as s	elected by the	11GSS bit of th	IE CM2CON1

2: TMR1GE bit must be set to use either T1G pin or C2OUT, as selected by the T1GSS bit of the CM2CON1 register, as a Timer1 gate source.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
CM2CON1	MC1OUT	MC2OUT	C1RSEL	C2RSEL	—	—	T1GSS	C2SYNC	000010	000010
INTCON	GIE	PEIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF	x000 000x	x000 0000x
PIE1	_	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	-000 0000	-000 0000
PIR1	-	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	-000 0000	-000 0000
TMR1H	Holding Re		xxxx xxxx	uuuu uuuu						
TMR1L	Holding Register for the Least Significant Byte of the 16-bit TMR1 Register									uuuu uuuu
T1CON	T1GINV	TMR1GE	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N	0000 0000	uuuu uuuu

TABLE 6-1: SUMMARY OF REGISTERS ASSOCIATED WITH TIMER1

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by the Timer1 module.

7.0 TIMER2 MODULE

The Timer2 module is an eight-bit timer with the following features:

- 8-bit timer register (TMR2)
- 8-bit period register (PR2)
- Interrupt on TMR2 match with PR2
- Software programmable prescaler (1:1, 1:4, 1:16)
- Software programmable postscaler (1:1 to 1:16)

See Figure 7-1 for a block diagram of Timer2.

7.1 Timer2 Operation

The clock input to the Timer2 module is the system instruction clock (Fosc/4). The clock is fed into the Timer2 prescaler, which has prescale options of 1:1, 1:4 or 1:16. The output of the prescaler is then used to increment the TMR2 register.

The values of TMR2 and PR2 are constantly compared to determine when they match. TMR2 will increment from 00h until it matches the value in PR2. When a match occurs, two things happen:

- TMR2 is reset to 00h on the next increment cycle
- The Timer2 postscaler is incremented

The match output of the Timer2/PR2 comparator is then fed into the Timer2 postscaler. The postscaler has postscale options of 1:1 to 1:16 inclusive. The output of the Timer2 postscaler is used to set the TMR2IF interrupt flag bit in the PIR1 register. The TMR2 and PR2 registers are both fully readable and writable. On any Reset, the TMR2 register is set to 00h and the PR2 register is set to FFh.

Timer2 is turned on by setting the TMR2ON bit in the T2CON register to a '1'. Timer2 is turned off by clearing the TMR2ON bit to a '0'.

The Timer2 prescaler is controlled by the T2CKPS bits in the T2CON register. The Timer2 postscaler is controlled by the TOUTPS bits in the T2CON register. The prescaler and postscaler counters are cleared when:

- A write to TMR2 occurs.
- A write to T2CON occurs.
- Any device Reset occurs (Power-on Reset, MCLR Reset, Watchdog Timer Reset, or Brown-out Reset).

Note:	TMR2	is	not	cleared	when	T2CON	is
	written.						

FIGURE 7-1: TIMER2 BLOCK DIAGRAM



U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0
bit 7							bit 0
Legend:							
R = Readable	e bit	W = Writable	bit	U = Unimplen	nented bit, rea	d as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown
bit 7	Unimplemen	ted: Read as '	0'				
bit 6-3	TOUTPS<3:0	>: Timer2 Out	out Postscaler	Select bits			
	0000 = 1:1 P	ostscaler					
	0001 = 1:2 P	ostscaler					
	0010 = 1:3 P	ostscaler					
	0011 = 1:4 P	ostscaler					
	0100 = 1.5 P	ostscaler					
	0101 = 1.0 P	ostscaler					
	0110 = 1.7 P	ostscaler					
	1000 = 1.9 P	ostscaler					
	1001 = 1:10 F	Postscaler					
	1010 = 1:11 F	Postscaler					
	1011 = 1:12 F	Postscaler					
	1100 = 1:13 H	Postscaler					
	1101 = 1:14	Postscaler					
	1110 = 1:15	Postscaler					
		Postscaler					
bit 2	TMR2ON: Tir	ner2 On bit					
	1 = Timer2 is	son					
	0 = Timer2 is	s off					
bit 1-0	T2CKPS<1:0	>: Timer2 Cloc	k Prescale Se	lect bits			
	00 = Prescale	er is 1					
	01 = Prescale	er is 4					
	1x = Prescale	er is 16					

REGISTER 7-1: T2CON: TIMER2 CONTROL REGISTER

TABLE 7-1: SUMMARY OF ASSOCIATED TIMER2 REGISTERS

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
INTCON	GIE	PEIE	TOIE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	x000 000x
PIE1	—	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	-000 0000	-000 0000
PIR1	—	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	-000 0000	-000 0000
PR2	Timer2 M	lodule Period	Register						1111 1111	1111 1111
TMR2	Holding Register for the 8-bit TMR2 Register									0000 0000
T2CON		TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0	-000 0000	-000 0000

Legend: x = unknown, u = unchanged, - = unimplemented read as '0'. Shaded cells are not used for Timer2 module.

8.0 COMPARATOR MODULE

Comparators are used to interface analog circuits to a digital circuit by comparing two analog voltages and providing a digital indication of their relative magnitudes. The comparators are very useful mixed signal building blocks because they provide analog functionality independent of the program execution. The analog Comparator module includes the following features:

- Independent comparator control
- Programmable input selection
- Comparator output is available internally/externally
- Programmable output polarity
- Interrupt-on-change
- · Wake-up from Sleep
- PWM shutdown
- Timer1 gate (count enable)
- Output synchronization to Timer1 clock input
- SR Latch
- Programmable and fixed voltage reference

Note: Only Comparator C2 can be linked to Timer1.

8.1 Comparator Overview

A single comparator is shown in Figure 8-1 along with the relationship between the analog input levels and the digital output. When the analog voltage at VIN+ is less than the analog voltage at VIN-, the output of the comparator is a digital low level. When the analog voltage at VIN+ is greater than the analog voltage at VIN-, the output of the comparator is a digital high level.



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8.2 Comparator Control

Each comparator has a separate control and configuration register: CM1CON0 for Comparator C1 and CM2CON0 for Comparator C2. In addition, Comparator C2 has a second control register, CM2CON1, for controlling the interaction with Timer1 and simultaneous reading of both comparator outputs.

The CM1CON0 and CM2CON0 registers (see Registers 8-1 and 8-2, respectively) contain the control and Status bits for the following:

- Enable
- Input selection
- Reference selection
- Output selection
- Output polarity

8.2.1 COMPARATOR ENABLE

Setting the CxON bit of the CMxCON0 register enables the comparator for operation. Clearing the CxON bit disables the comparator resulting in minimum current consumption.

8.2.2 COMPARATOR INPUT SELECTION

The CxCH<1:0> bits of the CMxCON0 register direct one of four analog input pins to the comparator inverting input.

Note:	To use CxIN+ and CxIN- pins as analog
	inputs, the appropriate bits must be set in
	the ANSEL and ANSELH registers and the
	corresponding TRIS bits must also be set
	to disable the output drivers.

8.2.3 COMPARATOR REFERENCE SELECTION

Setting the CxR bit of the CMxCON0 register directs an internal voltage reference or an analog input pin to the non-inverting input of the comparator. See **Section 8.10 "Comparator Voltage Reference"** for more information on the internal Voltage Reference module.

8.2.4 COMPARATOR OUTPUT SELECTION

The output of the comparator can be monitored by reading either the CxOUT bit of the CMxCON0 register or the MCxOUT bit of the CM2CON1 register. In order to make the output available for an external connection, the following conditions must be true:

- CxOE bit of the CMxCON0 register must be set
- Corresponding TRIS bit must be cleared
- CxON bit of the CMxCON0 register must be set

- Note 1: The CxOE bit overrides the PORT data latch. Setting the CxON has no impact on the port override.
 - 2: The internal output of the comparator is latched with each instruction cycle. Unless otherwise specified, external outputs are not latched.

8.2.5 COMPARATOR OUTPUT POLARITY

Inverting the output of the comparator is functionally equivalent to swapping the comparator inputs. The polarity of the comparator output can be inverted by setting the CxPOL bit of the CMxCON0 register. Clearing the CxPOL bit results in a non-inverted output.

Table 8-1 shows the output state versus input conditions, including polarity control.

TABLE 8-1: COMPARATOR OUTPUT STATE VS. INPUT CONDITIONS

Input Condition	CxPOL	CxOUT
CxVIN - > CxVIN +	0	0
CxVIN- < CxVIN+	0	1
CxVIN - > CxVIN +	1	1
CxVIN- < CxVIN+	1	0

8.3 Comparator Response Time

The comparator output is indeterminate for a period of time after the change of an input source or the selection of a new reference voltage. This period is referred to as the response time. The response time of the comparator differs from the settling time of the voltage reference. Therefore, both of these times must be considered when determining the total response time to a comparator input change. See the Comparator and Voltage Reference specifications in **Section 17.0 "Electrical Specifications"** for more details.

8.4 Comparator Interrupt Operation

The comparator interrupt flag can be set whenever there is a change in the output value of the comparator. Changes are recognized by means of a mismatch circuit which consists of two latches and an exclusiveor gate (see Figures 8-2 and 8-3). One latch is updated with the comparator output level when the CMxCON0 register is read. This latch retains the value until the next read of the CMxCON0 register or the occurrence of a Reset. The other latch of the mismatch circuit is updated on every Q1 system clock. A mismatch condition will occur when a comparator output change is clocked through the second latch on the Q1 clock cycle. At this point the two mismatch latches have opposite output levels which is detected by the exclusive-or gate and fed to the interrupt circuitry. The mismatch condition persists until either the CMxCON0 register is read or the comparator output returns to the previous state.

- Note 1: A write operation to the CMxCON0 register will also clear the mismatch condition because all writes include a read operation at the beginning of the write cycle.
 - **2:** Comparator interrupts will operate correctly regardless of the state of CxOE.

The comparator interrupt is set by the mismatch edge and not the mismatch level. This means that the interrupt flag can be reset without the additional step of reading or writing the CMxCON0 register to clear the mismatch registers. When the mismatch registers are cleared, an interrupt will occur upon the comparator's return to the previous state, otherwise no interrupt will be generated.

Software will need to maintain information about the status of the comparator output, as read from the CMxCON0 register, or CM2CON1 register, to determine the actual change that has occurred.

The CxIF bit of the PIR2 register is the comparator interrupt flag. This bit must be reset in software by clearing it to '0'. Since it is also possible to write a '1' to this register, an interrupt can be generated.

The CxIE bit of the PIE2 register and the PEIE and GIE bits of the INTCON register must all be set to enable comparator interrupts. If any of these bits are cleared, the interrupt is not enabled, although the CxIF bit of the PIR2 register will still be set if an interrupt condition occurs.

FIGURE 8-4: COMPARATOR INTERRUPT TIMING W/O CMxCON0 READ

FIGURE 8-5:

COMPARATOR INTERRUPT TIMING WITH CMxCON0 READ

reset by software



- Note 1: If a change in the CMxCON0 register (CxOUT) should occur when a read operation is being executed (start of the Q2 cycle), then the CxIF of the PIR2 register interrupt flag may not get set.
 - When either comparator is first enabled, bias circuitry in the Comparator module may cause an invalid output from the comparator until the bias circuitry is stable. Allow about 1 μs for bias settling then clear the mismatch condition and interrupt flags before enabling comparator interrupts.

8.5 Operation During Sleep

The comparator, if enabled before entering Sleep mode, remains active during Sleep. The additional current consumed by the comparator is shown separately in the **Section 17.0** "**Electrical Specifications**". If the comparator is not used to wake the device, power consumption can be minimized while in Sleep mode by turning off the comparator. Each comparator is turned off by clearing the CxON bit of the CMxCON0 register.

A change to the comparator output can wake-up the device from Sleep. To enable the comparator to wake the device from Sleep, the CxIE bit of the PIE2 register and the PEIE bit of the INTCON register must be set. The instruction following the Sleep instruction always executes following a wake from Sleep. If the GIE bit of the INTCON register is also set, the device will then execute the Interrupt Service Routine.

8.6 Effects of a Reset

A device Reset forces the CMxCON0 and CM2CON1 registers to their Reset states. This forces both comparators and the voltage references to their Off states.

R/W-0	R-0	R/W-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0	
C10N	C1OUT	C10E	C1POL	_	C1R	C1CH1	C1CH0	
bit 7							bit 0	
Legend:								
R = Reada	ıble bit	W = Writable	bit	U = Unimpler	mented bit, read	d as '0'		
-n = Value	at POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unki	nown	
bit 7	C1ON: Comp 1 = Compara 0 = Compara	arator C1 Ena tor C1 is enabl tor C1 is disab	ble bit ed led					
bit 6	bit 6 C1OUT: Comparator C1 Output bit $\frac{If C1POL = 1 \text{ (inverted polarity):}}{C1OUT = 0 \text{ when } C1VIN+ > C1VIN- C1OUT = 1 \text{ when } C1VIN+ < C1VIN- \frac{If C1POL = 0 \text{ (non-inverted polarity):}}{C1OUT = 1 \text{ when } C1VIN+ > C1VIN- C1OUT = 1 \text{ when } C1VIN+ > C1VIN-$							
bit 5	C1OE: Comp 1 = C1OUT is 0 = C1OUT is	arator C1 Outp present on th internal only	out Enable bit e C1OUT pin ⁽¹)				
bit 4	C1POL: Com 1 = C1OUT lo 0 = C1OUT lo	parator C1 Ou ogic is inverted	tput Polarity S rted	elect bit				
bit 3	Unimplemen	ted: Read as	0'					
bit 2	C1R: Compare 1 = C1VIN+ co 0 = C1VIN+ co	rator C1 Reference onnects to C1 onnects to C1	ence Select bit /REF output N+ pin	(non-inverting	input)			
bit 1-0	C1CH<1:0>: 00 = C12IN0 01 = C12IN1 10 = C12IN2 11 = C12IN3	Comparator C pin of C1 con pin of C1 con pin of C1 con pin of C1 con	1 Channel Sele nects to C1VIN nects to C1VIN nects to C1VIN nects to C1VIN	ect bit - - - -				
Note 1:	Comparator outpu	t requires the f	ollowing three	conditions: C1	OE = 1, C1ON	= 1 and corres	sponding port	

REGISTER 8-1: CM1CON0: COMPARATOR C1 CONTROL REGISTER 0

Note 1: Comparator output requires the following three conditions: C1OE = 1, C1ON = 1 and corresponding port TRIS bit = 0.

R/W-C) R-0	R/W-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0	
C2ON	I C2OUT	C2OE	C2POL	—	C2R	C2CH1	C2CH0	
bit 7							bit 0	
Legend:								
R = Read	able bit	W = Writable	bit	U = Unimpler	mented bit, rea	d as '0'		
-n = Value	e at POR	'1' = Bit is set	t	'0' = Bit is cle	ared	x = Bit is unkr	nown	
bit 7	C2ON: Comp 1 = Compara 0 = Compara	arator C2 Ena tor C2 is enab tor C2 is disab	ble bit led led					
bit 6	bit 6 $C2OUT: Comparator C2 Output bit$ $\frac{If C2POL = 1 (inverted polarity):}{C2OUT = 0 when C2VIN+ > C2VIN-C2OUT = 1 when C2VIN+ < C2VIN- If C2POL = 0 (non-inverted polarity): C2OUT = 1 when C2VIN+ > C2VIN-C2OUT = 1 when C2VIN+ > C2VIN+C2OUT = 1 when C2VIN+ > C2VIN+C2OUT = 1 when C2VIN+ > C2VIN+C2OUT = 1 when C2VIN+C2OUT$							
bit 5	C2OE: Comp 1 = C2OUT is 0 = C2OUT is	earator C2 Out s present on C s internal only	put Enable bit 2OUT pin ⁽¹⁾					
bit 4	C2POL: Com 1 = C2OUT lo 0 = C2OUT lo	parator C2 Ou ogic is inverted ogic is not inve	itput Polarity S l rted	elect bit				
bit 3	Unimplemen	ted: Read as	ʻ0'					
bit 2	bit 2 C2R: Comparator C2 Reference Select bits (non-inverting input) 1 = C2VIN+ connects to C2VREF 0 = C2VIN+ connects to C2IN+ pin							
bit 1-0	C2CH<1:0>: 00 = C12IN0- 01 = C12IN1- 10 = C12IN2- 11 = C12IN3-	Comparator C - pin of C2 con - pin of C2 con - pin of C2 con - pin of C2 con	2 Channel Sele nects to C2VIN nects to C2VIN nects to C2VIN nects to C2VIN	ect bits - - -				
Note 1:	Comparator outpu	t requires the	following three	conditions: C2	OE = 1. C2ON	l = 1 and corres	spondina port	

REGISTER 8-2: CM2CON0: COMPARATOR C2 CONTROL REGISTER 0

Note 1: Comparator output requires the following three conditions: C2OE = 1, C2ON = 1 and corresponding port TRIS bit = 0.

8.7 Analog Input Connection Considerations

A simplified circuit for an analog input is shown in Figure 8-6. Since the analog input pins share their connection with a digital input, they have reverse biased ESD protection diodes to VDD and VSS. The analog input, therefore, must be between VSS and VDD. If the input voltage deviates from this range by more than 0.6V in either direction, one of the diodes is forward biased and a latch-up may occur.

A maximum source impedance of $10 \text{ k}\Omega$ is recommended for the analog sources. Also, any external component connected to an analog input pin, such as a capacitor or a Zener diode, should have very little leakage current to minimize inaccuracies introduced.

- Note 1: When reading a PORT register, all pins configured as analog inputs will read as a '0'. Pins configured as digital inputs will convert as an analog input, according to the input specification.
 - 2: Analog levels on any pin defined as a digital input, may cause the input buffer to consume more current than is specified.



FIGURE 8-6: ANALOG INPUT MODEL

8.8 Additional Comparator Features

There are three additional comparator features:

- Timer1 count enable (gate)
- Synchronizing output with Timer1
- Simultaneous read of comparator outputs

8.8.1 COMPARATOR C2 GATING TIMER1

This feature can be used to time the duration or interval of analog events. Clearing the T1GSS bit of the CM2CON1 register will enable Timer1 to increment based on the output of Comparator C2. This requires that Timer1 is on and gating is enabled. See **Section 6.0 "Timer1 Module with Gate Control"** for details.

It is recommended to synchronize the comparator with Timer1 by setting the C2SYNC bit when the comparator is used as the Timer1 gate source. This ensures Timer1 does not miss an increment if the comparator changes during an increment.

8.8.2 SYNCHRONIZING COMPARATOR C2 OUTPUT TO TIMER1

The Comparator C2 output can be synchronized with Timer1 by setting the C2SYNC bit of the CM2CON1 register. When enabled, the C2 output is latched on the falling edge of the Timer1 clock source. If a prescaler is used with Timer1, the comparator output is latched after the prescaling function. To prevent a race condition, the comparator output is latched on the falling edge of the Timer1 clock source and Timer1 increments on the rising edge of its clock source. See the Comparator Block Diagram (Figures 8-2 and 8-3) and the Timer1 Block Diagram (Figure 6-1) for more information.

8.8.3 SIMULTANEOUS COMPARATOR OUTPUT READ

The MC1OUT and MC2OUT bits of the CM2CON1 register are mirror copies of both comparator outputs. The ability to read both outputs simultaneously from a single register eliminates the timing skew of reading separate registers.

Note 1: Obtaining the status of C1OUT or C2OUT by reading CM2CON1 does not affect the comparator interrupt mismatch registers.

REGISTER 8-3: CM2CON1: COMPARATOR C2 CONTROL REGISTER 1

R-0	R-0	R/W-0	R/W-0	U-0	U-0	R/W-1	R/W-0
MC1OUT	MC2OUT	C1RSEL	C2RSEL	—	—	T1GSS	C2SYNC
bit 7	•						bit 0

Legend:									
R = Readable	bit	W = Writable bit	U = Unimplemented bit,	read as '0'					
-n = Value at F	POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown					
bit 7	MC10UT: Mirror Copy of C10UT bit								
bit 6	MC2OUT: Mir	ror Copy of C2OUT bit							
bit 5	C1RSEL: Co	mparator C1 Reference	Select bit						
	1 = CVREF routed to C1VREF input of Comparator C1								
0 = Absolute voltage reference (0.6) routed to C1VREF input of Comparator C1 (or 1.2V precisi									
	reference	e on parts so equipped)							
bit 4	C2RSEL: Co	mparator C2 Reference	Select bit						
	1 = CVREF r	outed to C2VREF input	t of Comparator C2						
	0 = Absolute	voltage reference (0.6)) routed to C2VREF input of Co	omparator C2 (or 1.2V precision					
	reference	e on parts so equipped)							
bit 3-2	Unimplemen	ted: Read as '0'							
bit 1	T1GSS: Time	r1 Gate Source Select	bit						
	1 = Timer1 ga	ate source is T1G							
	0 = Timer1 ga	ate source is SYNCC2C	DUT.						
bit 0	C2SYNC: Co	mparator C2 Output Sy	nchronization bit						
	1 = Output is	synchronous to falling e	edge of Timer1 clock						
	0 = Output is	asynchronous							

8.9 Comparator SR Latch

The SR Latch module provides additional control of the comparator outputs. The module consists of a single SR latch and output multiplexers. The SR latch can be set, reset or toggled by the comparator outputs. The SR latch may also be set or reset, independent of comparator output, by control bits in the SRCON control register. The SR latch output multiplexers select whether the latch outputs or the comparator outputs are directed to the I/O port logic for eventual output to a pin.

8.9.1 LATCH OPERATION

The latch is a Set-Reset latch that does not depend on a clock source. Each of the Set and Reset inputs are active-high. Each latch input is connected to a comparator output and a software controlled pulse generator. The latch can be set by C1OUT or the PULSS bit of the SRCON register. The latch can be reset by C2OUT or the PULSR bit of the SRCON register. The latch is reset-dominant, therefore, if both Set and Reset inputs are high the latch will go to the Reset state. Both the PULSS and PULSR bits are self resetting which means that a single write to either of the bits is all that is necessary to complete a latch set or Reset operation.

8.9.2 LATCH OUTPUT

The SR<1:0> bits of the SRCON register control the latch output multiplexers and determine four possible output configurations. In these four configurations, the CxOUT I/O port logic is connected to:

- C1OUT and C2OUT
- C1OUT and SR latch Q
- · C2OUT and SR latch Q
- SR latch Q and \overline{Q}

After any Reset, the default output configuration is the unlatched C1OUT and C2OUT mode. This maintains compatibility with devices that do not have the SR latch feature.

The applicable TRIS bits of the corresponding ports must be cleared to enable the port pin output drivers. Additionally, the CxOE comparator output enable bits of the CMxCON0 registers must be set in order to make the comparator or latch outputs available on the output pins. The latch configuration enable states are completely independent of the enable states for the comparators.





R/W-0	R/W-0	R/W-0	R/W-0	R/S-0	R/S-0	U-0	R/W-0
SR1 ⁽²⁾) SR0 ⁽²⁾	C1SEN	C2REN	PULSS	PULSR		FVREN
bit 7		·					bit 0
Legend:				S = Bit is set	only -		
R = Reada	able bit	W = Writable	bit	U = Unimpler	nented bit, rea	d as '0'	
-n = Value	at POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown
bit 7	SR1: SR Late	ch Configuratio	n bit (2)				
	1 = C2OUT	Γ pin is the latcl	n <u>Q</u> output				
	0 = C2OUT	Γ pin is the C2 o	comparator ou	tput			
bit 6	SR0: SR Late	ch Configuratio	n bits ⁽²⁾				
	1 = C1OUT	Γ pin is the latcl	n Q output				
	0 = C1OUT	Γ pin is the C1 (Comparator ou	ıtput			
bit 5	C1SEN: C1 S	Set Enable bit					
	1 = C1 comp	parator output s	ets SR latch				
		barator output r	ias no effect ol	n SR latch			
bit 4	C2REN: C2	Reset Enable b					
	1 = C2 comp0 = C2 comp	parator output r parator output h	esets SR latch	n SR latch			
bit 3	PULSS: Puls	e the SET Inpu	ut of the SR La	tch bit			
	1 = Triggers	pulse generato	or to set SR lat	ch. Bit is imme	diately reset by	/ hardware.	
	0 = Does not	t trigger pulse g	generator				
bit 2	PULSR: Puls	se the Reset In	put of the SR L	_atch bit			
	1 = Triggers	pulse generato	or to reset SR I	atch. Bit is imm	nediately reset	by hardware.	
	0 = Does not	t trigger pulse o	generator				
bit 1	Unimplemen	nted: Read as '	0'				
bit 0	FVREN: Fixe	ed Voltage Refe	rence Enable	bit			
	1 = 0.6V Ref	ference FROM	INTOSC LDO	is enabled			
	0 = 0.6V Ref	terence FROM	INTOSC LDO	is disabled			
Note 1:	The CxOUT bit in	the CMxCON0	register will alv	ways reflect the	actual compara	ator output (not	the level on
	the pin), regardles	s of the SR late	h operation.				
2:	To enable an SR I configured.	Latch output to	the pin, the ap	propriate CxO	E and TRIS bit	s must be prop	erly

REGISTER 8-4: SRCON: SR LATCH CONTROL REGISTER

8.10 Comparator Voltage Reference

The Comparator Voltage Reference module provides an internally generated voltage reference for the comparators. The following features are available:

- Independent from Comparator operation
- Two 16-level voltage ranges
- Output clamped to Vss
- Ratiometric with VDD
- Fixed Reference (0.6V)

The VRCON register (Register 8-5) controls the Voltage Reference module shown in Figure 8-8.

The voltage source is selectable through both ends of the 16 connection resistor ladder network. Bit VRSS of the VRCON register selects either the internal or external voltage source.

The PIC16F883/884/886/887 allows the CVREF signal to be output to the RA2 pin of PORTA under certain configurations only. For more details, see Figure 8-9.

8.10.1 INDEPENDENT OPERATION

The comparator voltage reference is independent of the comparator configuration. Setting the VREN bit of the VRCON register will enable the voltage reference.

8.10.2 OUTPUT VOLTAGE SELECTION

The CVREF voltage reference has 2 ranges with 16 voltage levels in each range. Range selection is controlled by the VRR bit of the VRCON register. The 16 levels are set with the VR<3:0> bits of the VRCON register.

The CVREF output voltage is determined by the following equations:

EQUATION 8-1: CVREF OUTPUT VOLTAGE

VRR = 1 (low range): CVREF = (VR<3:0>/24) × VLADDER VRR = 0 (high range): CVREF = (VLADDER/4) + (VR<3:0> × VLADDER/32) VLADDER = VDD or ([VREF+] - [VREF-]) or VREF+

The full range of Vss to VDD cannot be realized due to the construction of the module. See Figure 8-8.

8.10.3 OUTPUT CLAMPED TO Vss

The CVREF output voltage can be set to Vss with no power consumption by configuring VRCON as follows:

- VREN = 0
- VRR = 1
- VR<3:0> = 0000

This allows the comparator to detect a zero-crossing while not consuming additional CVREF module current.

8.10.4 OUTPUT RATIOMETRIC TO VDD

The comparator voltage reference is VDD derived and therefore, the CVREF output changes with fluctuations in VDD. The tested absolute accuracy of the Comparator Voltage Reference can be found in **Section 17.0 "Electrical Specifications"**.

8.10.5 FIXED VOLTAGE REFERENCE

The fixed voltage reference is independent of VDD, with a nominal output voltage of 0.6V. This reference can be enabled by setting the FVREN bit of the SRCON register to '1'. This reference is always enabled when the HFINTOSC oscillator is active.

8.10.6 FIXED VOLTAGE REFERENCE STABILIZATION PERIOD

When the fixed Voltage Reference module is enabled, it will require some time for the reference and its amplifier circuits to stabilize. The user program must include a small delay routine to allow the module to settle. See the electrical specifications section for the minimum delay requirement.

8.10.7 VOLTAGE REFERENCE SELECTION

Multiplexers on the output of the Voltage Reference module enable selection of either the CVREF or fixed voltage reference for use by the comparators.

Setting the C1VREN bit of the VRCON register enables current to flow in the CVREF voltage divider and selects the CVREF voltage for use by C1. Clearing the C1VREN bit selects the fixed voltage for use by C1.

Setting the C2VREN bit of the VRCON register enables current to flow in the CVREF voltage divider and selects the CVREF voltage for use by C2. Clearing the C2VREN bit selects the fixed voltage for use by C2.

When both the C1VREN and C2VREN bits are cleared, current flow in the CVREF voltage divider is disabled minimizing the power drain of the voltage reference peripheral.







RA3	RA2	Comp. Reference (+)	Comp. Reference (-)	ADC Reference (+)	ADC Reference (-)	CFG1	CFG0	VRSS	VROE
I/O	I/O	AVdd	AVss	AVdd	AVss	0	0	0	0
I/O	CVREF	AVdd	AVss	AVdd	AVss	0	0	0	1
VREF+	VREF-	VREF+	VREF-	AVdd	AVss	0	0	1	0
VREF+	CVREF	VREF+	AVss	AVdd	AVss	0	0	1	1
VREF+	I/O	AVdd	AVss	Vref+	AVss	0	1	0	0
VREF+	CVREF	AVdd	AVss	Vref+	AVss	0	1	0	1
VREF+	VREF-	VREF+	VREF-	Vref+	AVss	0	1	1	0
VREF+	CVREF	VREF+	AVss	VREF+	AVss	0	1	1	1
I/O	VREF-	AVdd	AVss	AVdd	VREF-	1	0	0	0
I/O	VREF-	AVdd	AVss	AVdd	VREF-	1	0	0	1
VREF+	VREF-	VREF+	VREF-	AVdd	VREF-	1	0	1	0
VREF+	VREF-	VREF+	VREF-	AVDD	VREF-	1	0	1	1
VREF+	VREF-	AVdd	AVss	VREF+	VREF-	1	1	0	0
VREF+	VREF-	AVdd	AVss	Vref+	VREF-	1	1	0	1
VREF+	VREF-	VREF+	VREF-	VREF+	VREF-	1	1	1	0
VREF+	VREF-	VREF+	VREF-	VREF+	VREF-	1	1	1	1

TABLE 8-2: COMPARATOR AND ADC VOLTAGE REFERENCE PRIORITY

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
VREN	VROE	VRR	VRSS	VR3	VR2	VR1	VR0
bit 7	·						bit 0
Legend:							
R = Readable	bit	W = Writable	bit	U = Unimplei	mented bit, read	d as '0'	
-n = Value at F	POR	'1' = Bit is set		'0' = Bit is cle	eared	x = Bit is unki	nown
bit 7	VREN: Comp	arator C1 Volta	ige Reference	Enable bit			
	1 = CVREF ci	rcuit powered of	on				
	0 = CVREF ci	rcuit powered of	lown				
bit 6	VROE: Comp	parator C2 Volta	age Reference	Enable bit			
	1 = CVREF VO	oltage level is a	lso output on	the RA2/AN2/\	/REF-/CVREF/C2	2IN+ pin	
	0 = CVREF VC	oltage is discon	nected from th	ne RA2/AN2/V	REF-/CVREF/C2	N+ pin	
bit 5	VRR: CVREF	Range Selection	on bit				
	1 = Low range	e					
	0 = High rang	je					
bit 4	VRSS: Comp	arator VREF Ra	inge Selection	bit			
	1 = Compara	ator Reference	Source, CVRS	RC = (VREF+) -	(Vref-)		
	0 = Compara	ator Reference	Source, CVRS	RC = VDD - VSS	6		
bit 3-0	VR<3:0>: CV	REF Value Sele	ection $0 \leq VR <$	<3:0> ≤ 15			
	When VRR -		K <j:u>/24) ^ \ \n/1 ± (\/R~3(</j:u>	עט≀ 			
		$\underline{0}$. U WREF – VL	$\nu \nu \tau + \tau \nu \tau \sim 0.0$				

REGISTER 8-5: VRCON: VOLTAGE REFERENCE CONTROL REGISTER

TABLE 8-3:SUMMARY OF REGISTERS ASSOCIATED WITH THE COMPARATOR AND VOLTAGE
REFERENCE MODULES

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ANSEL	ANS7	ANS6	ANS5	ANS4	ANS3	ANS2	ANS1	ANS0	1111 1111	1111 1111
ANSELH	—		ANS13	ANS12	ANS11	ANS10	ANS9	ANS8	11 1111	11 1111
CM1CON0	C10N	C10UT	C10E	C1POL		C1R	C1CH1	C1CH0	0000 -000	0000 -000
CM2CON0	C2ON	C2OUT	C2OE	C2POL	_	C2R	C2CH1	C2CH0	0000 -000	0000 -000
CM2CON1	MC1OUT	MC2OUT	C1RSEL	C2RSEL		_	T1GSS	C2SYNC	000010	000010
INTCON	GIE	PEIE	TOIE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	0000 000x
PIE2	OSFIE	C2IE	C1IE	EEIE	BCLIE	ULPWUIE	—	CCP2IE	0000 00-0	0000 00-0
PIR2	OSFIF	C2IF	C1IF	EEIF	BCLIF	ULPWUIF	—	CCP2IF	0000 00-0	0000 00-0
PORTA	RA7	RA6	RA5	RA4	RA3	RA2	RA1	RA0	xxxx xxxx	uuuu uuuu
PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	xxxx xxxx	uuuu uuuu
SRCON	SR1	SR0	C1SEN	C2SEN	PULSS	PULSR	—	FVREN	0000 00-0	0000 00-0
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	1111 1111	1111 1111
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	1111 1111	1111 1111
VRCON	VREN	VROE	VRR	VRSS	VR3	VR2	VR1	VR0	0000 0000	0000 0000

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used for comparator.

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NOTES:

9.0 ANALOG-TO-DIGITAL CONVERTER (ADC) MODULE

The Analog-to-Digital Converter (ADC) allows conversion of an analog input signal to a 10-bit binary representation of that signal. This device uses analog inputs, which are multiplexed into a single sample and hold circuit. The output of the sample and hold is connected to the input of the converter. The converter generates a 10-bit binary result via successive approximation and stores the conversion result into the ADC result registers (ADRESL and ADRESH).

The ADC voltage reference is software selectable to either VDD or a voltage applied to the external reference pins.

The ADC can generate an interrupt upon completion of a conversion. This interrupt can be used to wake-up the device from Sleep.

Figure 9-1 shows the block diagram of the ADC.





9.1 ADC Configuration

When configuring and using the ADC the following functions must be considered:

- Port configuration
- Channel selection
- ADC voltage reference selection
- ADC conversion clock source
- Interrupt control
- Results formatting

9.1.1 PORT CONFIGURATION

The ADC can be used to convert both analog and digital signals. When converting analog signals, the I/O pin should be configured for analog by setting the associated TRIS and ANSEL bits. See the corresponding Port section for more information.

Note:	Analog voltages on any pin that is defined										
	as a digital input may cause the input										
	buffer to conduct excess current.										

9.1.2 CHANNEL SELECTION

The CHS bits of the ADCON0 register determine which channel is connected to the sample and hold circuit.

When changing channels, a delay is required before starting the next conversion. Refer to **Section 9.2 "ADC Operation"** for more information.

9.1.3 ADC VOLTAGE REFERENCE

The VCFG bits of the ADCON0 register provide independent control of the positive and negative voltage references. The positive voltage reference can be either VDD or an external voltage source. Likewise, the negative voltage reference can be either Vss or an external voltage source.

9.1.4 CONVERSION CLOCK

The source of the conversion clock is software selectable via the ADCS bits of the ADCON0 register. There are four possible clock options:

- Fosc/2
- Fosc/8
- Fosc/32
- FRC (dedicated internal oscillator)

The time to complete one bit conversion is defined as TAD. One full 10-bit conversion requires 11 TAD periods as shown in Figure 9-3.

For correct conversion, the appropriate TAD specification must be met. See A/D conversion requirements in **Section 17.0 "Electrical Specifications"** for more information. Table 9-1 gives examples of appropriate ADC clock selections.

Note: Unless using the FRC, any changes in the system clock frequency will change the ADC clock frequency, which may adversely affect the ADC result.

TABLE 9-1: ADC CLOCK PERIOD (TAD) Vs. DEVICE OPERATING FREQUENCIES (VDD \geq 3.0V)

ADC Clock	Period (TAD)				
ADC Clock Source	ADCS<2:0>	20 MHz	8 MHz	4 MHz	1 MHz
Fosc/2	000	100 ns (2)	250 ns ⁽²⁾	500 ns ⁽²⁾	2.0 μs
Fosc/8	001	400 ns ⁽²⁾	1.0 μs (2)	2.0 μs	8.0 μs ⁽³⁾
Fosc/32	010	1.6 μs	4.0 μs	8.0 μs ⁽³⁾	32.0 μs ⁽³⁾
FRC	x11	2-6 μs ^(1,4)	2-6 μs ^(1,4)	2-6 μs ^(1,4)	2-6 μs ^(1,4)

Legend: Shaded cells are outside of recommended range.

Note 1: The FRC source has a typical TAD time of $4 \mu s$ for VDD > 3.0V.

2: These values violate the minimum required TAD time.

3: For faster conversion times, the selection of another clock source is recommended.

4: When the device frequency is greater than 1 MHz, the FRC clock source is only recommended if the conversion will be performed during Sleep.

FIGURE 9-2: ANALOG-TO-DIGITAL CONVERSION TAD CYCLES

TCY to TAD TAD1	TAD2	TAD3	TAD4	TAD5	TAD6	TAD7	TAD8	TAD9	TAD10	TAD11	
↑ ↑ <u>↑</u>	b9	b8	b7	b6	b5	b4	b3	b2	b1	b0	
Conver	sion St	arts									
Holding Capa	acitor is	s Disco	nnecte	d from	Analog	g Input	(typica	lly 100	ns)		
Set GO/DONE	bit					ADRE GO bi	SH an	d ADR ared,	ESL re	gisters] are loaded,
						ADIF Holdiı	bit is s ng capa	et, acitor is	s conne	ected to	analog input

9.1.5 INTERRUPTS

The ADC module allows for the ability to generate an interrupt upon completion of an Analog-to-Digital conversion. The ADC interrupt flag is the ADIF bit in the PIR1 register. The ADC interrupt enable is the ADIE bit in the PIE1 register. The ADIF bit must be cleared in software.

Note:	The ADIF bit is set at the completion of
	every conversion, regardless of whether
	or not the ADC interrupt is enabled.

This interrupt can be generated while the device is operating or while in Sleep. If the device is in Sleep, the interrupt will wake-up the device. Upon waking from Sleep, the next instruction following the SLEEP instruction is always executed. If the user is attempting to wake-up from Sleep and resume in-line code execution, the global interrupt must be disabled. If the global interrupt is enabled, execution will switch to the Interrupt Service Routine.

Please see **Section 14.3** "Interrupts" for more information.

9.1.6 RESULT FORMATTING

The 10-bit A/D conversion result can be supplied in two formats, left justified or right justified. The ADFM bit of the ADCON0 register controls the output format.

Figure 9-3 shows the two output formats.





9.2 ADC Operation

9.2.1 STARTING A CONVERSION

To enable the ADC module, the ADON bit of the ADCON0 register must be set to a '1'. Setting the GO/DONE bit of the ADCON0 register to a '1' will start the Analog-to-Digital conversion.

Note:	The GO/DONE bit should not be set in the							
	same instruction that turns on the ADC.							
	Refer to Section 9.2.6 "A/D Conversion							
	Procedure".							

9.2.2 COMPLETION OF A CONVERSION

When the conversion is complete, the ADC module will:

- Clear the GO/DONE bit
- Set the ADIF flag bit
- Update the ADRESH:ADRESL registers with new conversion result

9.2.3 TERMINATING A CONVERSION

If a conversion must be terminated before completion, the GO/DONE bit can be cleared in software. The ADRESH:ADRESL registers will not be updated with the partially complete Analog-to-Digital conversion sample. Instead, the ADRESH:ADRESL register pair will retain the value of the previous conversion. Additionally, a 2 TAD delay is required before another acquisition can be initiated. Following this delay, an input acquisition is automatically started on the selected channel.

Note:	A device Reset forces all registers to their									
	Reset state. Thus, the ADC module is									
	turned off and any pending conversion is terminated.									

9.2.4 ADC OPERATION DURING SLEEP

The ADC module can operate during Sleep. This requires the ADC clock source to be set to the FRC option. When the FRC clock source is selected, the ADC waits one additional instruction before starting the conversion. This allows the SLEEP instruction to be executed, which can reduce system noise during the conversion. If the ADC interrupt is enabled, the device will wake-up from Sleep when the conversion completes. If the ADC interrupt is disabled, the ADC module is turned off after the conversion completes, although the ADON bit remains set.

When the ADC clock source is something other than FRC, a SLEEP instruction causes the present conversion to be aborted and the ADC module is turned off, although the ADON bit remains set.

9.2.5 SPECIAL EVENT TRIGGER

The ECCP Special Event Trigger allows periodic ADC measurements without software intervention. When this trigger occurs, the GO/DONE bit is set by hardware and the Timer1 counter resets to zero.

Using the Special Event Trigger does not assure proper ADC timing. It is the user's responsibility to ensure that the ADC timing requirements are met.

See Section 11.0 "Capture/Compare/PWM Modules (CCP1 and CCP2)" for more information.

9.2.6 A/D CONVERSION PROCEDURE

This is an example procedure for using the ADC to perform an Analog-to-Digital conversion:

- 1. Configure Port:
 - Disable pin output driver (See TRIS register)
 - Configure pin as analog
- 2. Configure the ADC module:
 - Select ADC conversion clock
 - Configure voltage reference
 - Select ADC input channel
 - Select result format
 - Turn on ADC module
- 3. Configure ADC interrupt (optional):
 - Clear ADC interrupt flag
 - Enable ADC interrupt
 - Enable peripheral interrupt
 - Enable global interrupt⁽¹⁾
- 4. Wait the required acquisition time⁽²⁾.
- 5. Start conversion by setting the GO/DONE bit.
- 6. Wait for ADC conversion to complete by one of the following:
 - Polling the GO/DONE bit
 - Waiting for the ADC interrupt (interrupts enabled)
- 7. Read ADC Result
- 8. Clear the ADC interrupt flag (required if interrupt is enabled).
 - **Note 1:** The global interrupt can be disabled if the user is attempting to wake-up from Sleep and resume in-line code execution.
 - 2: See Section 9.3 "A/D Acquisition Requirements".

EXAMPLE 9-1: A/D CONVERSION

;This code block configures the ADC ; for polling, Vdd and Vss as reference, Frc clock and AN0 input. ;Conversion start & polling for completion ; are included. BANKSEL ADCON1 B'10000000' ;right justify MOVIW MOVWF ADCON1 ;Vdd and Vss as Vref BANKSEL TRISA ; BSF TRISA,0 ;Set RA0 to input BANKSEL ANSEL ANSEL,0 BSF ;Set RA0 to analog ADCON0 BANKSEL ; B'11000001' ;ADC Frc clock, MOVLW ;AN0, On MOVWF ADCON0 CALL SampleTime ;Acquisiton delay ADCON0,GO ;Start conversion BSF BTFSC ADCON0,GO ;Is conversion done? GOTO ;No, test again \$-1 BANKSEL ADRESH ; MOVF ADRESH,W ;Read upper 2 bits MOVWF ;store in GPR space RESULTHI BANKSEL ADRESL : ADRESL,W MOVF ;Read lower 8 bits MOVWF RESULTLO ;Store in GPR space

9.2.7 ADC REGISTER DEFINITIONS

The following registers are used to control the operation of the ADC.

Note:	For ANSEL and ANSELH registers, see
	Register 3-3 and Register 3-4, respectively.

REGISTER 9-1: ADCON0: A/D CONTROL REGISTER 0

ADCS<1:0>: A/D Conversion Clock Select bits

R/W-0	R/W-0						
ADCS1	ADCS0	CHS3	CHS2	CHS1	CHS0	GO/DONE	ADON
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

00 = Fosc/201 = Fosc/810 = Fosc/32 11 = FRC (clock derived from a dedicated internal oscillator = 500 kHz max) bit 5-2 CHS<3:0>: Analog Channel Select bits 0000 = ANO 0001 = AN1 0010 = AN2 0011 = AN3 0100 = AN4 0101 = AN5 0110 = AN6 0111 = AN7 1000 **= AN8** 1001 = AN9 1010 = AN10 1011 = AN11 1100 = AN12 1101 = AN13 1110 = CVREF 1111 = Fixed Ref (0.6 volt fixed reference) bit 1 GO/DONE: A/D Conversion Status bit 1 = A/D conversion cycle in progress. Setting this bit starts an A/D conversion cycle. This bit is automatically cleared by hardware when the A/D conversion has completed. 0 = A/D conversion completed/not in progress bit 0 ADON: ADC Enable bit 1 = ADC is enabled 0 = ADC is disabled and consumes no operating current

bit 7-6

R/W-0	U-0	R/W-0	R/W-0	U-0	U-0	U-0	U-0	
ADFM		VCFG1	VCFG0		—	—	—	
bit 7							bit 0	
Legend:								
R = Readable	bit	W = Writable	bit	U = Unimpler	mented bit, read	d as '0'		
-n = Value at F	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown	
bit 7	ADFM: A/D C	Conversion Res	ult Format Sel	lect bit				
	1 = Right just 0 = Left justifi	ified ied						
bit 6	Unimplemen	ted: Read as '	0'					
bit 5	VCFG1: Voltage Reference bit							
	1 = VREF- pin							
	0 = VSS							
bit 4	VCFG0: Volta	age Reference	bit					
	1 = VREF+ pir 0 = VDD	1						
bit 3-0	Unimplemen	ted: Read as '	0'					

REGISTER 9-2: ADCON1: A/D CONTROL REGISTER 1

'1' = Bit is set

R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	
ADRES9	ADRES8	ADRES7	ADRES6	ADRES5	ADRES4	ADRES3	ADRES2	
bit 7							bit 0	
Legend:								
R = Readable bit W = Writable bit			t	U = Unimplemented bit, read as '0'				

'0' = Bit is cleared

x = Bit is unknown

REGISTER 9-3: ADRESH: ADC RESULT REGISTER HIGH (ADRESH) ADFM = 0

ADRES<9:2>: ADC Result Register bits bit 7-0 Upper 8 bits of 10-bit conversion result

-n = Value at POR

REGISTER 9-4: ADRESL: ADC RESULT REGISTER LOW (ADRESL) ADFM = 0

R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
ADRES1	ADRES0	—	—	—	—	—	-
bit 7							bit 0
Legend:							
R = Readable b	it	W = Writable bi	it	U = Unimplemented bit, read as '0'			
-n = Value at PC	= Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is un			x = Bit is unknow	wn		

bit 7-6	ADRES<1:0>: ADC Result Register bits
	Lower 2 bits of 10-bit conversion result

bit 5-0 Reserved: Do not use.

REGISTER 9-5: ADRESH: ADC RESULT REGISTER HIGH (ADRESH) ADFM = 1

R/W-x	R/W-x						
—	—	—	—	—	—	ADRES9	ADRES8
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as	s 'O'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-2 Reserved: Do not use.

bit 1-0 ADRES<9:8>: ADC Result Register bits Upper 2 bits of 10-bit conversion result

REGISTER 9-6: ADRESL: ADC RESULT REGISTER LOW (ADRESL) ADFM = 1

| R/W-x |
|--------|--------|--------|--------|--------|--------|--------|--------|
| ADRES7 | ADRES6 | ADRES5 | ADRES4 | ADRES3 | ADRES2 | ADRES1 | ADRES0 |
| bit 7 | | | | | | | bit 0 |

Legend:				
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown	

bit 7-0 ADRES<7:0>: ADC Result Register bits Lower 8 bits of 10-bit conversion result
9.3 A/D Acquisition Requirements

For the ADC to meet its specified accuracy, the charge holding capacitor (CHOLD) must be allowed to fully charge to the input channel voltage level. The Analog Input model is shown in Figure 9-4. The source impedance (Rs) and the internal sampling switch (Rss) impedance directly affect the time required to charge the capacitor CHOLD. The sampling switch (Rss) impedance varies over the device voltage (VDD), see Figure 9-4. **The maximum recommended impedance for analog sources is 10 k** Ω . As the source impedance is decreased, the acquisition time may be decreased. After the analog input channel is selected (or changed),

an A/D acquisition must be done before the conversion can be started. To calculate the minimum acquisition time, Equation 9-1 may be used. This equation assumes that 1/2 LSb error is used (1024 steps for the ADC). The 1/2 LSb error is the maximum error allowed for the ADC to meet its specified resolution.

EQUATION 9-1: ACQUISITION TIME EXAMPLE

Assumptions: Temperature = 50°C and external impedance of 10k
$$\Omega$$
 5.0V VDD

$$TACQ = Amplifier Settling Time + Hold Capacitor Charging Time + Temperature Coefficient
= TAMP + TC + TCOFF
= 2 \mu s + TC + [(Temperature - 25°C)(0.05 \mu s/°C)]
The value for TC can be approximated with the following equations:
$$V_{APPLIED}\left(1 - \frac{1}{2047}\right) = V_{CHOLD} \qquad :[1] VCHOLD charged to within 1/2 lsb
$$V_{APPLIED}\left(1 - e^{\frac{-TC}{RC}}\right) = V_{CHOLD} \qquad :[2] VCHOLD charge response to VAPPLIED
$$V_{APPLIED}\left(1 - e^{\frac{-TC}{RC}}\right) = V_{CHOLD} \qquad :[2] VCHOLD charge response to VAPPLIED
$$V_{APPLIED}\left(1 - e^{\frac{-TC}{RC}}\right) = V_{APPLIED}\left(1 - \frac{1}{2047}\right) \qquad :combining [1] and [2]$$
Solving for TC:

$$T_{C} = -CHOLD(RIC + RSS + RS) \ln(1/2047)$$

$$= -10pF(1k\Omega + 7k\Omega + 10k\Omega) \ln(0.0004885)$$

$$= 1.37\mu s$$
Therefore:

$$T_{ACQ} = 2\mu s + 1.37\mu s + [(50°C - 25°C)(0.05\mu s/°C)]$$$$$$$$$$

Note 1: The reference voltage (VREF) has no effect on the equation, since it cancels itself out.

2: The charge holding capacitor (CHOLD) is not discharged after each conversion.

3: The maximum recommended impedance for analog sources is $10 \text{ k}\Omega$. This is required to meet the pin leakage specification.

 $= 4.67 \mu S$

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Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
ADCON0	ADCS1	ADCS0	CHS3	CHS2	CHS1	CHS0	GO/DONE	ADON	0000 0000	0000 0000
ADCON1	ADFM	_	VCFG1	VCFG0	_	_	_	_	0-00	-000
ANSEL	ANS7	ANS6	ANS5	ANS4	ANS3	ANS2	ANS1	ANS0	1111 1111	1111 1111
ANSELH	—		ANS13	ANS12	ANS11	ANS10	ANS9	ANS8	11 1111	11 1111
ADRESH	A/D Result	Register High	n Byte						XXXX XXXX	uuuu uuuu
ADRESL	A/D Result	Register Low	Byte						XXXX XXXX	uuuu uuuu
INTCON	GIE	PEIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	0000 000x
PIE1	_	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	-000 0000	-000 0000
PIR1	_	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	-000 0000	-000 0000
PORTA	RA7	RA6	RA5	RA4	RA3	RA2	RA1	RA0	XXXX XXXX	uuuu uuuu
PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	XXXX XXXX	uuuu uuuu
PORTE	-	_	_	_	RE3	RE2	RE1	RE0	xxxx	uuuu
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	1111 1111	1111 1111
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	1111 1111	1111 111
TRISE	_	_	_	_	TRISE3	TRISE2	TRISE1	TRISE0	1111	111

TABLE 9-2: SUMMARY OF ASSOCIATED ADC REGISTERS

Legend: x = unknown, u = unchanged, - = unimplemented read as '0'. Shaded cells are not used for ADC module.

NOTES:

10.0 DATA EEPROM AND FLASH PROGRAM MEMORY CONTROL

The Data EEPROM and Flash program memory are readable and writable during normal operation (full VDD range). These memories are not directly mapped in the register file space. Instead, they are indirectly addressed through the Special Function Registers (SFRs). There are six SFRs used to access these memories:

- EECON1
- EECON2
- EEDAT
- EEDATH
- EEADR
- EEADRH (bit 4 on PIC16F886/PIC16F887 only)

When interfacing the data memory block, EEDAT holds the 8-bit data for read/write, and EEADR holds the address of the EEDAT location being accessed. These devices have 256 bytes of data EEPROM with an address range from 0h to 0FFh.

When accessing the program memory block of the PIC16F886/PIC16F887 devices, the EEDAT and EEDATH registers form a 2-byte word that holds the 14-bit data for read/write, and the EEADR and EEADRH registers form a 2-byte word that holds the 12-bit address of the EEPROM location being read. The PIC16F882 devices have 2K words of program EEPROM with an address range from 0h to 07FFh. The PIC16F883/PIC16F884 devices have 4K words of program EEPROM with an address range from 0h to 0FFFh. The program memory allows one-word reads.

The EEPROM data memory allows byte read and write. A byte write automatically erases the location and writes the new data (erase before write).

The write time is controlled by an on-chip timer. The write/erase voltages are generated by an on-chip charge pump rated to operate over the voltage range of the device for byte or word operations.

Depending on the setting of the Flash Program Memory Self Write Enable bits WRT<1:0> of the Configuration Word Register 2, the device may or may not be able to write certain blocks of the program memory. However, reads from the program memory are allowed.

When the device is code-protected, the CPU may continue to read and write the data EEPROM memory and Flash program memory. When code-protected, the device programmer can no longer access data or program memory.

10.1 EEADR and EEADRH Registers

The EEADR and EEADRH registers can address up to a maximum of 256 bytes of data EEPROM or up to a maximum of 8K words of program EEPROM.

When selecting a program address value, the MSB of the address is written to the EEADRH register and the LSB is written to the EEADR register. When selecting a data address value, only the LSB of the address is written to the EEADR register.

10.1.1 EECON1 AND EECON2 REGISTERS

EECON1 is the control register for EE memory accesses.

Control bit EEPGD determines if the access will be a program or data memory access. When clear, as it is when reset, any subsequent operations will operate on the data memory. When set, any subsequent operations will operate on the program memory. Program memory can only be read.

Control bits RD and WR initiate read and write, respectively. These bits cannot be cleared, only set, in software. They are cleared in hardware at completion of the read or write operation. The inability to clear the WR bit in software prevents the accidental, premature termination of a write operation.

The WREN bit, when set, will allow a write operation to data EEPROM. On power-up, the WREN bit is clear. The WRERR bit is set when a write operation is interrupted by a MCLR or a WDT Time-out Reset during normal operation. In these situations, following Reset, the user can check the WRERR bit and rewrite the location.

Interrupt flag bit EEIF of the PIR2 register is set when write is complete. It must be cleared in the software.

EECON2 is not a physical register. Reading EECON2 will read all '0's. The EECON2 register is used exclusively in the data EEPROM write sequence.

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
EEDAT7	EEDAT6	EEDAT5	EEDAT4	EEDAT3	EEDAT2	EEDAT1	EEDAT0
bit 7							bit 0
Legend:							
R = Readable bit	:	W = Writable bi	t	U = Unimplem	ented bit, read as	'0'	
-n = Value at POR '1' = Bit is set				'0' = Bit is cleared x = Bit is unknown			

REGISTER 10-1: EEDAT: EEPROM DATA REGISTER

bit 7-0

-n = Value at POR

EEDAT<7:0>: 8 Least Significant Address bits to Write to or Read from data EEPROM or Read from program memory

'0' = Bit is cleared

REGISTER 10-2: EEADR: EEPROM ADDRESS REGISTER

EEADD7	EEADBE		EEADD4	EEADB2			EEADDO
EEADRI	EEADRO	EEADRS	EEADK4	EEADKS	EEADKZ	EEADKI	EEADRU
bit 7							bit 0
Legend:							
R = Readable bit W = Writable bit		t	U = Unimplemented bit, read as '0'				

bit 7-0 **EEADR<7:0>**: 8 Least Significant Address bits for EEPROM Read/Write Operation⁽¹⁾ or Read from program memory

REGISTER 10-3: EEDATH: EEPROM DATA HIGH BYTE REGISTER

'1' = Bit is set

U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	—	EEDATH5	EEDATH4	EEDATH3	EEDATH2	EEDATH1	EEDATH0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as	ʻ0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6 Unimplemented: Read as '0'

bit 5-0 **EEDATH<5:0>:** 6 Most Significant Data bits from program memory

REGISTER 10-4: EEADRH: EEPROM ADDRESS HIGH BYTE REGISTER

	00	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
_	_	EEADRH4 ⁽¹⁾	EEADRH3	EEADRH2	EEADRH1	EEADRH0
						bit 0
	W = Writable bi	t	U = Unimpleme	ented bit, read as	'0'	
	'1' = Bit is set		'0' = Bit is clear	red	x = Bit is unknow	wn
_		W = Writable bi '1' = Bit is set	W = Writable bit '1' = Bit is set	— EEADRH4 ⁽¹⁾ EEADRH3 W = Writable bit U = Unimplement '1' = Bit is set '0' = Bit is clear	— EEADRH4 ⁽¹⁾ EEADRH3 EEADRH2 W = Writable bit U = Unimplemented bit, read as '1' = Bit is set '0' = Bit is cleared	— EEADRH4 ⁽¹⁾ EEADRH3 EEADRH2 EEADRH1 W = Writable bit U = Unimplemented bit, read as '0' x = Bit is unkno '1' = Bit is set '0' = Bit is cleared x = Bit is unkno

bit 7-5 Unimplemented: Read as '0'

bit 4-0 EEADRH<4:0>: Specifies the 4 Most Significant Address bits or high bits for program memory reads

Note 1: PIC16F886/PIC16F887 only.

x = Bit is unknown

REGISTER 10-5: EECON1: EEPROM CONTROL REGISTER

R/W-x	U-0	U-0	U-0	R/W-x	R/W-0	R/S-0	R/S-0
EEPGD	—	—	—	WRERR	WREN	WR	RD
bit 7							bit 0
Legend:							
S = Bit can only	y be set						
R = Readable	bit	W = Writable	bit	U = Unimpler	mented bit, read	l as '0'	
-n = Value at P	OR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	iown
bit 7	EEPGD: Prog	jram/Data EEP	ROM Select	bit			
	1 = Accesses 0 = Accesses	s program men s data memory	nory				
bit 6-4		ted: Read as ')'				
bit 3	WRERR: EEF	PROM Error Fla	ag bit				
	1 = A write o	peration is prer	naturely tern	ninated (any MC	LR Reset, any	WDT Reset du	ring
	normal of	peration or BO	R Reset)				
1	0 = I he write	operation com	pleted				
bit 2	WREN: EEPF	ROM Write Ena	ble bit				
	1 = Allows wr $0 = Inhibits wr$	rite cycles	FFPROM				
bit 1	WR: Write Co	ntrol bit					
	1 = Initiates a	write cycle (TI	ne bit is cleai	red by hardware	once write is co	omplete. The W	/R bit can only
	be set, no	ot cleared, in se	oftware.)				
	0 = Write cyc	le to the data E	EPROM is a	complete			
bit 0	RD: Read Co	ntrol bit					
	1 = Initiates a	a memory read	d (the RD is	cleared in har	dware and can	only be set, r	ot cleared, in
	0 = Does not	, initiate a mem	ory read				
			-				

10.1.2 READING THE DATA EEPROM MEMORY

To read a data memory location, the user must write the address to the EEADR register, clear the EEPGD control bit of the EECON1 register, and then set control bit RD. The data is available at the very next cycle, in the EEDAT register; therefore, it can be read in the next instruction. EEDAT will hold this value until another read or until it is written to by the user (during a write operation).

FXAMPLE 10-1	ΔΑΤΑ FEPROM READ

BANKSEL	EEADR	;
MOVLW	DATA_EE_ADDR	;
MOVWF	EEADR	;Data Memory
		;Address to read
BANKSEL	EECON1	;
BCF	EECON1, EEPGD	; Point to DATA memory
BSF	EECON1, RD	;EE Read
BANKSEL	EEDAT	;
MOVF	EEDAT, W	;W = EEDAT
BCF	STATUS, RP1	;Bank 0

10.1.3 WRITING TO THE DATA EEPROM MEMORY

To write an EEPROM data location, the user must first write the address to the EEADR register and the data to the EEDAT register. Then the user must follow a specific sequence to initiate the write for each byte.

The write will not initiate if the above sequence is not followed exactly (write 55h to EECON2, write AAh to EECON2, then set WR bit) for each byte. Interrupts should be disabled during this code segment.

Additionally, the WREN bit in EECON1 must be set to enable write. This mechanism prevents accidental writes to data EEPROM due to errant (unexpected) code execution (i.e., lost programs). The user should keep the WREN bit clear at all times, except when updating EEPROM. The WREN bit is not cleared by hardware.

After a write sequence has been initiated, clearing the WREN bit will not affect this write cycle. The WR bit will be inhibited from being set unless the WREN bit is set.

At the completion of the write cycle, the WR bit is cleared in hardware and the EE Write Complete Interrupt Flag bit (EEIF) is set. The user can either enable this interrupt or poll this bit. EEIF must be cleared by software.

	BANKSEL	EEADR		i
	MOVLW	DATA_EE_A	ADDR	i
	MOVWF	EEADR		;Data Memory Address to write
	MOVLW	DATA_EE_D	DATA	;
	MOVWF	EEDAT		;Data Memory Value to write
	BANKSEL	EECON1		;
	BCF	EECON1, E	EEPGD	;Point to DATA memory
	BSF	EECON1, W	WREN	;Enable writes
	DCE	TNECON	777	
	BCF	INTCON, G	31E 375	; DISADLE INIS.
	BIFSC	INTCON, G	JΙΕ	;SEE AN576
	GOTO	Ş-2		
	MOVLW	55h		i
be'	MOVWF	EECON2		;Write 55h
ner	MOVLW	AAh		i
Red	MOVWF	EECON2		;Write AAh
- 0)	BSF	EECON1, W	WR	;Set WR bit to begin write
	BSF	INTCON, G	GIE	;Enable INTs.
	STEED			Whit for interrupt to gianal write complete
	DCE	EECON1 W	זאים סיג	, Wait for interrupt to signal write complete
	BCF	EECONI, W	NKEN	park 0
	BCF	STATUS, R	KPU DDa	;BARK U
	BCF	STATUS, R	KLT	

EXAMPLE 10-2: DATA EEPROM WRITE

10.1.4 READING THE FLASH PROGRAM MEMORY

To read a program memory location, the user must write the Least and Most Significant address bits to the EEADR and EEADRH registers, set the EEPGD control bit of the EECON1 register, and then set control bit RD. Once the read control bit is set, the program memory Flash controller will use the second instruction cycle to read the data. This causes the second instruction immediately following the "BSF EECON1, RD" instruction to be ignored. The data is available in the very next cycle, in the EEDAT and EEDATH registers; therefore, it can be read as two bytes in the following instructions.

EXAMPLE 10-3: FLASH PROGRAM READ

EEDAT and EEDATH registers will hold this value until another read or until it is written to by the user.

- Note 1: The two instructions following a program memory read are required to be NOPS. This prevents the user from executing a two-cycle instruction on the next instruction after the RD bit is set.
 - If the WR bit is set when EEPGD = 1, it will be immediately reset to '0' and no operation will take place.

	BANKSE	L EEADR	;
	MOVLW	MS_PROG_EE_ADDR	;
	MOVWF	EEADRH	;MS Byte of Program Address to read
	MOVLW	LS_PROG_EE_ADDR	;
	MOVWF	EEADR	;LS Byte of Program Address to read
	BANKSE	L EECON1	;
	BSF	EECON1, EEPGD	;Point to PROGRAM memory
ъe	BSF	EECON1, RD	;EE Read
Require Sequenc	NOP		;First instruction after BSF EECON1,RD executes normally
	NOP		;Any instructions here are ignored as program
· · · ·			;memory is read in second cycle after BSF EECON1,RD
;			
	BANKSE	L EEDAT	;
	MOVF	EEDAT, W	;W = LS Byte of Program Memory
	MOVWF	LOWPMBYTE	;
	MOVF	EEDATH, W	;W = MS Byte of Program EEDAT
	MOVWF	HIGHPMBYTE	;
	BCF	STATUS, RP1	;Bank 0

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FIGURE 10-1: FLASH PROGRAM MEMORY READ CYCLE EXECUTION

10.2 Writing to Flash Program Memory

Flash program memory may only be written to if the destination address is in a segment of memory that is not write-protected, as defined in bits WRT<1:0> of the Configuration Word Register 2. Flash program memory must be written in eight-word blocks (four-word blocks for 4K memory devices). See Figures 10-2 and 10-3 for more details. A block consists of eight words with sequential addresses, with a lower boundary defined by an address, where EEADR<2:0> = 000. All block writes to program memory are done as 16-word erase by eight-word write operations. The write operation is edge-aligned and cannot occur across boundaries.

To write program data, it must first be loaded into the buffer registers (see Figure 10-2). This is accomplished by first writing the destination address to EEADR and EEADRH and then writing the data to EEDATA and EEDATH. After the address and data have been set up, then the following sequence of events must be executed:

- 1. Set the EEPGD control bit of the EECON1 register.
- 2. Write 55h, then AAh, to EECON2 (Flash programming sequence).
- 3. Set the WR control bit of the EECON1 register.

All eight buffer register locations should be written to with correct data. If less than eight words are being written to in the block of eight words, then a read from the program memory location(s) not being written to must be performed. This takes the data from the program location(s) not being written and loads it into the EEDATA and EEDATH registers. Then the sequence of events to transfer data to the buffer registers must be executed.

To transfer data from the buffer registers to the program memory, the EEADR and EEADRH must point to the last location in the eight-word block (EEADR<2:0> = 111). Then the following sequence of events must be executed:

- 1. Set the EEPGD control bit of the EECON1 register.
- 2. Write 55h, then AAh, to EECON2 (Flash programming sequence).
- 3. Set control bit WR of the EECON1 register to begin the write operation.

The user must follow the same specific sequence to initiate the write for each word in the program block, writing each program word in sequence (000, 001, 010, 011, 100, 101, 110, 111). When the write is performed on the last word (EEADR<2:0> = 111), a block of sixteen words is automatically erased and the content of the eight word buffer registers are written into the program memory.

After the "BSF EECON1, WR" instruction, the processor requires two cycles to set up the erase/write operation. The user must place two NOP instructions after the WR bit is set. Since data is being written to buffer registers, the writing of the first seven words of the block appears to occur immediately. The processor will halt internal operations for the typical 4 ms, only during the cycle in which the erase takes place (i.e., the last word of the sixteen-word block erase). This is not Sleep mode as the clocks and peripherals will continue to run. After the eight-word write cycle, the processor will resume operation with the third instruction after the EECON1 write instruction. The above sequence must be repeated for the higher eight words.

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An example of the complete eight-word write sequence is shown in Example 10-4. The initial address is loaded into the EEADRH and EEADR register pair; the eight words of data are loaded using indirect addressing.

EXAMPLE 10-4: WRITING TO FLASH PROGRAM MEMORY

; '	This	write r	coutine assumes the	following:					
;	;								
;	1. A	valid s	starting address (t	he least significant bits = `00')is loaded in ADDRH:ADDRL					
;	2. Tł	ne 8 byt	es of data are loa	ded, starting at the address in DATADDR					
;	3. AI	DDRH, AI	DRL and DATADDR ar	e all located in shared data memory 0x70 - 0x7f					
;									
		BSF	STATUS, RP1						
		BCF	STATUS, RPO	; Bank 2					
		MOVF	ADDRH,W	; Load initial address					
		MOVWF	EEADRH	;					
		MOVF	ADDRL,W	;					
		MOVWF	EEADR	;					
		MOVF	DATAADDR,W	; Load initial data address					
		MOVWF	FSR	;					
LO	ЭР	MOVF	INDF,W	; Load first data byte into lower					
		MOVWF	EEDATA	;					
		INCF	FSR,F	; Next byte					
		MOVF	INDF,W	; Load second data byte into upper					
		MOVWF	EEDATH	i					
		INCF	FSR,F	;					
		BSF	STATUS, RPO	; Bank 3					
		BSF	EECON1, EEPGD	; Point to program memory					
		BSF	EECON1,WREN	; Enable writes					
		BCF	INTCON,GIE	; Disable interrupts (if using)					
		BTFSC	INTCON,GIE	; See AN576					
		GOTO	1-2						
	— 0	MOVLW	55h	; Start of required write sequence:					
	ired ance	MOVWF	EECON2	; Write 55h					
	aue	MOVLW	AAh	;					
	SeR	MOVWF	EECON2	; Write AAh					
		BSF	EECON1,WR	; Set WR bit to begin write					
		NOP		; Any instructions here are ignored as processor					
				; halts to begin write sequence					
		NOP		; processor will stop here and wait for write complete					
				; after write processor continues with 3rd instruction					
		BCF	EECON1,WREN	; Disable writes					
		BSF	INTCON,GIE	; Enable interrupts (if using)					
		BCF	STATUS, RPO	; Bank 2					
		INCF	EEADR, F	; Increment address					
		MOVF	EEADR,W	; Check if lower two bits of address are `00'					
		ANDLW	0x07	; Indicates when four words have been programmed					
		XORLW	0x07	;					
		BTFSC	STATUS, Z	; Exit if more than eight words,					
		GOTO	LOOP	; Continue if less than eight words					

10.3 Write Verify

Depending on the application, good programming practice may dictate that the value written to the data EEPROM should be verified (see Example 10-5) to the desired value to be written.

EXAMPLE 10-5: WRITE VERIFY

BANKSEL	EEDAT	;
MOVF	EEDAT, W	;EEDAT not changed
		;from previous write
BANKSEL	EECON1	;
BSF	EECON1, RD	;YES, Read the
		;value written
BANKSEL	EEDAT	;
XORWF	EEDAT, W	;
BTFSS	STATUS, Z	;Is data the same
GOTO	WRITE_ERR	;No, handle error
:		;Yes, continue
BCF	STATUS, RP1	;Bank 0

10.3.1 USING THE DATA EEPROM

The data EEPROM is a high-endurance, byte addressable array that has been optimized for the storage of frequently changing information (e.g., program variables or other data that are updated often). When variables in one section change frequently, while variables in another section do not change, it is possible to exceed the total number of write cycles to the EEPROM (specification D124) without exceeding the total number of write cycles to a single byte (specifications D120 and D120A). If this is the case, then a refresh of the array must be performed. For this reason, variables that change infrequently (such as constants, IDs, calibration, etc.) should be stored in Flash program memory.

10.4 Protection Against Spurious Write

There are conditions when the user may not want to write to the data EEPROM memory. To protect against spurious EEPROM writes, various mechanisms have been built in. On power-up, WREN is cleared. Also, the Power-up Timer (64 ms duration) prevents EEPROM write.

The write initiate sequence and the WREN bit together help prevent an accidental write during:

- Brown-out
- Power Glitch
- Software Malfunction

10.5 Data EEPROM Operation During Code-Protect

Data memory can be code-protected by programming the CPD bit in the Configuration Word Register 1 (Register 14-1) to '0'. When the data memory is code-protected, only the CPU is able to read and write data to the data EEPROM. It is recommended to code-protect the program memory when code-protecting data memory. This prevents anyone from programming zeroes over the existing code (which will execute as NOPS) to reach an added routine, programmed in unused program memory, which outputs the contents of data memory. Programming unused locations in program memory to '0' will also help prevent data memory code protection from becoming breached.

TABLE 10-1: SUMMARY OF REGISTERS ASSOCIATED WITH DATA EEPROM

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
EECON1	EEPGD	—	—	—	WRERR	WREN	WR	RD	x x000	0 q000
EECON2	EEPROM C	ontrol Regis	ster 2 (not a p	physical registe	er)					
EEADR	EEADR7	EEADR6	EEADR5	EEADR4	EEADR3	EEADR2	EEADR1	EEADR0	0000 0000	0000 0000
EEADRH	_	—	—	EEADRH4 ⁽¹⁾	EEADRH3	EEADRH2	EEADRH1	EEADRH0	0 0000	0 0000
EEDAT	EEDAT7	EEDAT6	EEDAT5	EEDAT4	EEDAT3	EEDAT2	EEDAT1	EEDAT0	0000 0000	0000 0000
EEDATH	_	_	EEDATH5	EEDATH4	EEDATH3	EEDATH2	EEDATH1	EEDATH0	00 0000	00 0000
INTCON	GIE	PEIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	0000 000x
PIE2	OSFIE	C2IE	C1IE	EEIE	BCLIE	ULPWUIE	_	CCP2IE	0000 00-0	0000 00-0
PIR2	OSFIF	C2IF	C1IF	EEIF	BCLIF	ULPWUIF	—	CCP2IF	0000 00-0	0000 00-0

 $\label{eq:legend: Legend: Legend: x = unknown, u = unchanged, -= unimplemented read as `0', q = value depends upon condition. Shaded cells are not used by data EEPROM module.$

Note 1: PIC16F886/PIC16F887 only.

NOTES:

11.0 CAPTURE/COMPARE/PWM MODULES (CCP1 AND CCP2)

This device contains one Enhanced Capture/Compare/ PWM (CCP1) and Capture/Compare/PWM module (CCP2). The CCP1 and CCP2 modules are identical in operation, with the exception of the Enhanced PWM features available on CCP1 only. See **Section 11.6** "**PWM (Enhanced Mode)**" for more information.

Note: CCPRx and CCPx throughout this document refer to CCPR1 or CCPR2 and CCP1 or CCP2, respectively.

11.1 Enhanced Capture/Compare/PWM (CCP1)

The Enhanced Capture/Compare/PWM module is a peripheral which allows the user to time and control different events. In Capture mode, the peripheral allows the timing of the duration of an event. The Compare mode allows the user to trigger an external event when a predetermined amount of time has expired. The PWM mode can generate a Pulse-Width Modulated signal of varying frequency and duty cycle.

Table 11-1 shows the timer resources required by the ECCP module.

TABLE 11-1: ECCP MODE – TIMER RESOURCES REQUIRED

ECCP Mode	Timer Resource
Capture	Timer1
Compare	Timer1
PWM	Timer2

REGISTER 11-1: CCP1CON: ENHANCED CCP1 CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
P1M1	P1M0	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0
bit 7							bit 0
Legend:							
R = Readable bi	t	W = Writable bi	t	U = Unimpleme	ented bit, read as	s '0'	
-n = Value at PC	R	'1' = Bit is set		'0' = Bit is clear	red	x = Bit is unkno	own
bit 7-6	P1M<1:0>: PV If CCP1M<3:2: xx = P1A as: If CCP1M<3:2: 00 = Single 0 01 = Full-Brid 10 = Half-Brid 11 = Full-Brid	VM Output Config > = 00, 01, 10; signed as Capture > = 11; output; P1A modu dge output forwar dge output; P1A, dge output revers	uration bits e/Compare inpu lated; P1B, P10 d; P1D modula P1B modulated e; P1B modulated	it; P1B, P1C, P1E C, P1D assigned ted; P1A active; F with dead-band c ted; P1C active; F	D assigned as po as port pins P1B, P1C inactiv control; P1C, P1E P1A, P1D inactiv	ort pins e Dassigned as po e	rt pins
bit 5-4	bit 5-4 DC1B<1:0>: PWM Duty Cycle Least Significant bits Capture mode: Unused. Compare mode: Unused. PWM mode: These bits are the two LSbs of the PWM duty cycle. The eight MSbs are found in CCPR11						
bit 3-0	CCP1M<3:0>: 0000 = Captu 0001 = Unus 0010 = Comp 0011 = Unus 0100 = Captu 0101 = Captu 0110 = Captu 0110 = Captu 1001 = Comp 1001 = Comp 1010 = Comp 1010 = Comp 1010 = Comp 1010 = PWM 1101 = PWM 1111 = PWM	ECCP Mode Sel ure/Compare/PW ed (reserved) pare mode, toggle ed (reserved) ure mode, every f ure mode, every f pare mode, every 4 ure mode, every 4 pare mode, set out pare mode, gener pare mode, trigge mode; P1A, P1C mode; P1A, P1C mode; P1A, P1C	ect bits M off (resets EC e output on mate ising edge ith rising edge th rising edge th rising edge toput on match output on match output on match acte software infor r special event cactive-high; P cactive-high; P1 cactive-low; P1 cactive-low; P1	CCP module) ch (CCP1IF bit is set h (CCP1IF bit is set rerrupt on match ((CCP1IF bit is se 1B, P1D active-hi 1B, P1D active-lo B, P1D active-lov	set) et) (CCP1IF bit is se t; CCP1 resets T gh w yh	et, CCP1 pin is u MR1 or TMR2	naffected)

11.2 Capture/Compare/PWM (CCP2)

The Capture/Compare/PWM module is a peripheral which allows the user to time and control different events. In Capture mode, the peripheral allows the timing of the duration of an event. The Compare mode allows the user to trigger an external event when a predetermined amount of time has expired. The PWM mode can generate a Pulse-Width Modulated signal of varying frequency and duty cycle.

The timer resources used by the module are shown in Table 11-2.

Additional information on CCP modules is available in the Application Note AN594, *"Using the CCP Modules"* (DS00594).

TABLE 11-2: CCP MODE – TIMER RESOURCES REQUIRED

CCP Mode	Timer Resource
Capture	Timer1
Compare	Timer1
PWM	Timer2

REGISTER 11-2: CCP2CON: CCP2 CONTROL REGISTER

U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
	—	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0
bit 7				· · · · · · · · · · · · · · · · · · ·			bit 0
Legend:							
R = Readable	bit	W = Writable	bit	U = Unimplen	nented bit, read	d as '0'	
-n = Value at P	OR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown
bit 7-6 bit 5-4	Unimplement DC2B<1:0>: I Capture mode Unused. Compare mode Unused. PWM mode: These bits are	ted: Read as ' PWM Duty Cyc <u>2:</u> de: e the two LSbs	o' cle Least Signi of the PWM d	ficant bits uty cycle. The e	eight MSbs are	e found in CCP	R2L.
DIT 3-U	 Contained of a mode celections Contained of a mode celection of a mo					'2 pin D	

11.3 Capture Mode

In Capture mode, the CCPRxH, CCPRxL register pair captures the 16-bit value of the TMR1 register when an event occurs on pin CCPx. An event is defined as one of the following and is configured by the CCP1M<3:0> bits of the CCP1CON register:

- Every falling edge
- Every rising edge
- Every 4th rising edge
- Every 16th rising edge

When a capture is made, the Interrupt Request Flag bit CCPxIF of the PIRx register is set. The interrupt flag must be cleared in software. If another capture occurs before the value in the CCPRxH, CCPRxL register pair is read, the old captured value is overwritten by the new captured value (see Figure 11-1).

11.3.1 CCP PIN CONFIGURATION

In Capture mode, the CCPx pin should be configured as an input by setting the associated TRIS control bit.

Note:	If the CCPx pin is configured as an output,
	a write to the port can cause a capture
	condition.

FIGURE 11-1: CAPTURE MODE OPERATION BLOCK DIAGRAM



11.3.2 TIMER1 MODE SELECTION

Timer1 must be running in Timer mode or Synchronized Counter mode for the CCP module to use the capture feature. In Asynchronous Counter mode, the capture operation may not work.

11.3.3 SOFTWARE INTERRUPT

When the Capture mode is changed, a false capture interrupt may be generated. The user should keep the CCPxIE interrupt enable bit of the PIEx register clear to avoid false interrupts. Additionally, the user should clear the CCPxIF interrupt flag bit of the PIRx register following any change in Operating mode.

11.3.4 CCP PRESCALER

There are four prescaler settings specified by the CCPxM<3:0> bits of the CCPxCON register. Whenever the CCP module is turned off, or the CCP module is not in Capture mode, the prescaler counter is cleared. Any Reset will clear the prescaler counter.

Switching from one capture prescaler to another does not clear the prescaler and may generate a false interrupt. To avoid this unexpected operation, turn the module off by clearing the CCPxCON register before changing the prescaler (see Example 11-1).

EXAMPLE 11-1: CHANGING BETWEEN CAPTURE PRESCALERS

BANKSEL	CCP1CON	;Set Bank bits to point
		;to CCP1CON
CLRF	CCP1CON	;Turn CCP module off
MOVLW	NEW_CAPT_PS	;Load the W reg with
		; the new prescaler
		; move value and CCP ON
MOVWF	CCP1CON	;Load CCP1CON with this
		; value

11.4 Compare Mode

In Compare mode, the 16-bit CCPRx register value is constantly compared against the TMR1 register pair value. When a match occurs, the CCPx module may:

- Toggle the CCPx output
- Set the CCPx output
- Clear the CCPx output
- Generate a Special Event Trigger
- Generate a Software Interrupt

The action on the pin is based on the value of the CCPxM<3:0> control bits of the CCPxICON register.

All Compare modes can generate an interrupt.





- Clear TMR1H and TMR1L registers.
- NOT set interrupt flag bit TMR1IF of the PIR1 register.
- Set the GO/DONE bit to start the ADC conversion.

11.4.1 CCP PIN CONFIGURATION

The user must configure the CCPx pin as an output by clearing the associated TRIS bit.

Note:	Clearing the CCP1CON register will force				
	the CCPx compare output latch to the				
	default low level. This is not the PORT I/O				
	data latch.				

11.4.2 TIMER1 MODE SELECTION

In Compare mode, Timer1 must be running in either Timer mode or Synchronized Counter mode. The compare operation may not work in Asynchronous Counter mode.

11.4.3 SOFTWARE INTERRUPT MODE

When Generate Software Interrupt mode is chosen (CCPxM<3:0> = 1010), the CCPx module does not assert control of the CCPx pin (see the CCP1CON register).

11.4.4 SPECIAL EVENT TRIGGER

When Special Event Trigger mode is chosen (CCPxM<3:0> = 1011), the CCPx module does the following:

- Resets Timer1
- Starts an ADC conversion if ADC is enabled

The CCPx module does not assert control of the CCPx pin in this mode (see the CCPxCON register).

The Special Event Trigger output of the CCP occurs immediately upon a match between the TMR1H, TMR1L register pair and the CCPRxH, CCPRxL register pair. The TMR1H, TMR1L register pair is not reset until the next rising edge of the Timer1 clock. This allows the CCPRxH, CCPRxL register pair to effectively provide a 16-bit programmable period register for Timer1.

Note 1: The Special Event Trigger from the CCP module does not set interrupt flag bit TMRxIF of the PIR1 register.

2: Removing the match condition by changing the contents of the CCPRxH and CCPRxL register pair, between the clock edge that generates the Special Event Trigger and the clock edge that generates the Timer1 Reset, will preclude the Reset from occurring.

11.5 PWM Mode

The PWM mode generates a Pulse-Width Modulated signal on the CCPx pin. The duty cycle, period and resolution are determined by the following registers:

- PR2
- T2CON
- CCPRxL
- CCPxCON

In Pulse-Width Modulation (PWM) mode, the CCP module produces up to a 10-bit resolution PWM output on the CCPx pin. Since the CCPx pin is multiplexed with the PORT data latch, the TRIS for that pin must be cleared to enable the CCPx pin output driver.

Note:	Clearing	the	CCPxCON	register	will
	relinquish	CCP	x control of th	ne CCPx p	oin.

Figure 11-3 shows a simplified block diagram of PWM operation.

Figure 11-4 shows a typical waveform of the PWM signal.

For a step-by-step procedure on how to set up the CCP module for PWM operation, see **Section 11.5.7 "Setup for PWM Operation"**.

FIGURE 11-3: SIMPLIFIED PWM BLOCK DIAGRAM



2: In PWM mode, CCPRxH is a read-only register.

The PWM output (Figure 11-4) has a time base (period) and a time that the output stays high (duty cycle).

FIGURE 11-4: CCP PWM OUTPUT



11.5.1 PWM PERIOD

The PWM period is specified by the PR2 register of Timer2. The PWM period can be calculated using the formula of Equation 11-1.

EQUATION 11-1: PWM PERIOD

 $PWM Period = [(PR2) + 1] \bullet 4 \bullet Tosc \bullet$ (TMR2 Prescale Value)

When TMR2 is equal to PR2, the following three events occur on the next increment cycle:

- TMR2 is cleared
- The CCPx pin is set. (Exception: If the PWM duty cycle = 0%, the pin will not be set.)
- The PWM duty cycle is latched from CCPRxL into CCPRxH.

Note:	The Timer2 postscaler (see Section 7.1
	"Timer2 Operation") is not used in the
	determination of the PWM frequency.

11.5.2 PWM DUTY CYCLE

The PWM duty cycle is specified by writing a 10-bit value to multiple registers: CCPRxL register and DCxB<1:0> bits of the CCPxCON register. The CCPRxL contains the eight MSbs and the DCxB<1:0> bits of the CCPxCON register contain the two LSbs. CCPRxL and DCxB<1:0> bits of the CCPxCON register can be written to at any time. The duty cycle value is not latched into CCPRxH until after the period completes (i.e., a match between PR2 and TMR2 registers occurs). While using the PWM, the CCPRxH register is read-only.

Equation 11-2 is used to calculate the PWM pulse width.

Equation 11-3 is used to calculate the PWM duty cycle ratio.

EQUATION 11-2: PULSE WIDTH

$$Pulse Width = (CCPRxL:CCPxCON < 5:4>) \bullet$$

TOSC • (TMR2 Prescale Value)

EQUATION 11-3: DUTY CYCLE RATIO

Duty Cycle Ratio = $\frac{(CCPRxL:CCPxCON < 5:4>)}{4(PR2 + 1)}$

The CCPRxH register and a 2-bit internal latch are used to double buffer the PWM duty cycle. This double buffering is essential for glitchless PWM operation.

The 8-bit timer TMR2 register is concatenated with either the 2-bit internal system clock (Fosc), or 2 bits of the prescaler, to create the 10-bit time base. The system clock is used if the Timer2 prescaler is set to 1:1.

When the 10-bit time base matches the CCPRxH and 2-bit latch, then the CCPx pin is cleared (see Figure 11-3).

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11.5.3 PWM RESOLUTION

The resolution determines the number of available duty cycles for a given period. For example, a 10-bit resolution will result in 1024 discrete duty cycles, whereas an 8-bit resolution will result in 256 discrete duty cycles.

The maximum PWM resolution is 10 bits when PR2 is 255. The resolution is a function of the PR2 register value as shown by Equation 11-4.

EQUATION 11-4: PWM RESOLUTION

Resolution =
$$\frac{\log[4(PR2 + 1)]}{\log(2)}$$
 bits

Note: If the pulse width value is greater than the period the assigned PWM pin(s) will remain unchanged.

TABLE 11-3:	EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (Fosc = 20 MHz)
-------------	---

PWM Frequency	1.22 kHz	4.88 kHz	19.53 kHz	78.12 kHz	156.3 kHz	208.3 kHz
Timer Prescale (1, 4, 16)	16	4	1	1	1	1
PR2 Value	0xFF	0xFF	0xFF	0x3F	0x1F	0x17
Maximum Resolution (bits)	10	10	10	8	7	6.6

TABLE 11-4: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (Fosc = 8 MHz)

PWM Frequency	1.22 kHz	4.90 kHz	19.61 kHz	76.92 kHz	153.85 kHz	200.0 kHz
Timer Prescale (1, 4, 16)	16	4	1	1	1	1
PR2 Value	0x65	0x65	0x65	0x19	0x0C	0x09
Maximum Resolution (bits)	8	8	8	6	5	5

11.5.4 OPERATION IN SLEEP MODE

In Sleep mode, the TMR2 register will not increment and the state of the module will not change. If the CCPx pin is driving a value, it will continue to drive that value. When the device wakes up, TMR2 will continue from its previous state.

11.5.5 CHANGES IN SYSTEM CLOCK FREQUENCY

The PWM frequency is derived from the system clock frequency. Any changes in the system clock frequency will result in changes to the PWM frequency. See Section 4.0 "Oscillator Module (With Fail-Safe Clock Monitor)" for additional details.

11.5.6 EFFECTS OF RESET

Any Reset will force all ports to Input mode and the CCP registers to their Reset states.

11.5.7 SETUP FOR PWM OPERATION

The following steps should be taken when configuring the CCP module for PWM operation:

- 1. Disable the PWM pin (CCPx) output drivers as an input by setting the associated TRIS bit.
- 2. Set the PWM period by loading the PR2 register.
- 3. Configure the CCP module for the PWM mode by loading the CCPxCON register with the appropriate values.
- Set the PWM duty cycle by loading the CCPRxL register and DCxB<1:0> bits of the CCPxCON register.
- 5. Configure and start Timer2:
 - Clear the TMR2IF interrupt flag bit of the PIR1 register.
 - Set the Timer2 prescale value by loading the T2CKPS bits of the T2CON register.
 - Enable Timer2 by setting the TMR2ON bit of the T2CON register.
- 6. Enable PWM output after a new PWM cycle has started:
 - Wait until Timer2 overflows (TMR2IF bit of the PIR1 register is set).
 - Enable the CCPx pin output driver by clearing the associated TRIS bit.

11.6 PWM (Enhanced Mode)

The Enhanced PWM Mode can generate a PWM signal on up to four different output pins with up to 10-bits of resolution. It can do this through four different PWM output modes:

- Single PWM
- Half-Bridge PWM
- Full-Bridge PWM, Forward mode
- Full-Bridge PWM, Reverse mode

To select an Enhanced PWM mode, the P1M bits of the CCP1CON register must be set appropriately.

Note: The PWM Enhanced mode is available on the Enhanced Capture/Compare/PWM module (CCP1) only.

The PWM outputs are multiplexed with I/O pins and are designated P1A, P1B, P1C and P1D. The polarity of the PWM pins is configurable and is selected by setting the CCP1M bits in the CCP1CON register appropriately.

Table 11-5 shows the pin assignments for each Enhanced PWM mode.

Figure 11-5 shows an example of a simplified block diagram of the Enhanced PWM module.

Note: To prevent the generation of an incomplete waveform when the PWM is first enabled, the ECCP module waits until the start of a new PWM period before generating a PWM signal.





Note 1: The TRIS register value for each PWM output must be configured appropriately.

2: Clearing the CCPxCON register will relinquish ECCP control of all PWM output pins.

3: Any pin not used by an Enhanced PWM mode is available for alternate pin functions.

TABLE 11-5: EXAMPLE PIN ASSIGNMENTS FOR VARIOUS PWM ENHANCED MODES

ECCP Mode	P1M<1:0>	CCP1/P1A	P1B	P1C	P1D
Single	0.0	Yes ⁽¹⁾	Yes ⁽¹⁾	Yes ⁽¹⁾	Yes ⁽¹⁾
Half-Bridge	10	Yes	Yes	No	No
Full-Bridge, Forward	01	Yes	Yes	Yes	Yes
Full-Bridge, Reverse	11	Yes	Yes	Yes	Yes

Note 1: Pulse Steering enables outputs in Single mode.

FIGURE 11-6: EXAMPLE PWM (ENHANCED MODE) OUTPUT RELATIONSHIPS (ACTIVE-HIGH STATE)

			1			1
00	(Single Output)	P1A Modulated	 Delay	_v (1)	Delay(1)	
		P1A Modulated				
10	(Half-Bridge)	P1B Modulated	i			
		P1A Active				
01	(Full-Bridge,	P1B Inactive			1 1 1	
01	Forward)	P1C Inactive			1 1 	
		P1D Modulated				
		P1A Inactive	_ ;		1 1 1	<u> </u>
11	(Full-Bridge,	P1B Modulated				
	Reverse)	P1C Active	_ <u>;</u>			;
		P1D Inactive			1 1 1	<u> </u>
Rela	tionships: • Period = 4 * Tos • Pulse Width = To	c * (PR2 + 1) * (TMR2 Pre osc * (CCPR1L<7:0>:CCF	escale Value) P1CON<5:4>) *	(TMR2 Prescal	e Value)	

0 0	(Single Output)	P1A Modulated	i			
		P1A Modulated				
10	(Half-Bridge)	P1B Modulated	Dela	ay(''		
(Full-Bridge, ⁰¹ Forward)	P1A Active			· · ·		
	P1B Inactive	<u>:</u>		<u> </u>	1 1 1	
	P1C Inactive	_ :		 	1 1 1	
	P1D Modulated					
11 (Full-Bridge, Reverse)	P1A Inactive			1 1 1		
	P1B Modulated _	=			1 1 1	
	P1C Active _					
	P1D Inactive					

FIGURE 11-7: EXAMPLE ENHANCED PWM OUTPUT RELATIONSHIPS (ACTIVE-LOW STATE)

ote 1: Dead-band delay is programmed using the PWM1CON register (Section 11.6.6 "Programmable Dead-Band Delay Mode").

11.6.1 HALF-BRIDGE MODE

In Half-Bridge mode, two pins are used as outputs to drive push-pull loads. The PWM output signal is output on the CCPx/P1A pin, while the complementary PWM output signal is output on the P1B pin (see Figure 11-9). This mode can be used for Half-Bridge applications, as shown in Figure 11-9, or for Full-Bridge applications, where four power switches are being modulated with two PWM signals.

In Half-Bridge mode, the programmable dead-band delay can be used to prevent shoot-through current in Half-Bridge power devices. The value of the PDC<6:0> bits of the PWM1CON register sets the number of instruction cycles before the output is driven active. If the value is greater than the duty cycle, the corresponding output remains inactive during the entire cycle. See **Section 11.6.6 "Programmable Dead-Band Delay Mode"** for more details of the dead-band delay operations. Since the P1A and P1B outputs are multiplexed with the PORT data latches, the associated TRIS bits must be cleared to configure P1A and P1B as outputs.

FIGURE 11-8: EXAMPLE OF HALF-BRIDGE PWM OUTPUT



FIGURE 11-9: EXAMPLE OF HALF-BRIDGE APPLICATIONS



11.6.2 FULL-BRIDGE MODE

In Full-Bridge mode, all four pins are used as outputs. An example of Full-Bridge application is shown in Figure 11-10.

In the Forward mode, pin CCP1/P1A is driven to its active state, pin P1D is modulated, while P1B and P1C will be driven to their inactive state as shown in Figure 11-11.

In the Reverse mode, P1C is driven to its active state, pin P1B is modulated, while P1A and P1D will be driven to their inactive state as shown Figure 11-11. P1A, P1B, P1C and P1D outputs are multiplexed with the PORT data latches. The associated TRIS bits must be cleared to configure the P1A, P1B, P1C and P1D pins as outputs.

FIGURE 11-10: EXAMPLE OF FULL-BRIDGE APPLICATION



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FIGURE 11-11: **EXAMPLE OF FULL-BRIDGE PWM OUTPUT**

11.6.2.1 Direction Change in Full-Bridge Mode

In the Full-Bridge mode, the P1M1 bit in the CCP1CON register allows users to control the forward/reverse direction. When the application firmware changes this direction control bit, the module will change to the new direction on the next PWM cycle.

A direction change is initiated in software by changing the P1M1 bit of the CCP1CON register. The following sequence occurs four Timer2 cycles prior to the end of the current PWM period:

- The modulated outputs (P1B and P1D) are placed in their inactive state.
- The associated unmodulated outputs (P1A and P1C) are switched to drive in the opposite direction.
- PWM modulation resumes at the beginning of the next period.

See Figure 11-12 for an illustration of this sequence.

The Full-Bridge mode does not provide dead-band delay. As one output is modulated at a time, dead-band delay is generally not required. There is a situation where dead-band delay is required. This situation occurs when both of the following conditions are true:

- 1. The direction of the PWM output changes when the duty cycle of the output is at or near 100%.
- 2. The turn off time of the power switch, including the power device and driver circuit, is greater than the turn on time.

Figure 11-13 shows an example of the PWM direction changing from forward to reverse, at a near 100% duty cycle. In this example, at time t1, the output P1A and P1D become inactive, while output P1C becomes active. Since the turn off time of the power devices is longer than the turn on time, a shoot-through current will flow through power devices QC and QD (see Figure 11-10) for the duration of 't'. The same phenomenon will occur to power devices QA and QB for PWM direction change from reverse to forward.

If changing PWM direction at high duty cycle is required for an application, two possible solutions for eliminating the shoot-through current are:

- 1. Reduce PWM duty cycle for one PWM period before changing directions.
- 2. Use switch drivers that can drive the switches off faster than they can drive them on.

Other options to prevent shoot-through current may exist.

FIGURE 11-12: EXAMPLE OF PWM DIRECTION CHANGE



- **Note 1:** The direction bit P1M1 of the CCP1CON register is written any time during the PWM cycle.
 - 2: When changing directions, the P1A and P1C signals switch before the end of the current PWM cycle. The modulated P1B and P1D signals are inactive at this time. The length of this time is four Timer2 counts.



FIGURE 11-13: EXAMPLE OF PWM DIRECTION CHANGE AT NEAR 100% DUTY CYCLE

11.6.3 START-UP CONSIDERATIONS

When any PWM mode is used, the application hardware must use the proper external pull-up and/or pull-down resistors on the PWM output pins.

Note: When the microcontroller is released from Reset, all of the I/O pins are in the high-impedance state. The external circuits must keep the power switch devices in the Off state until the microcontroller drives the I/O pins with the proper signal levels or activates the PWM output(s).

The CCP1M<1:0> bits of the CCP1CON register allow the user to choose whether the PWM output signals are active-high or active-low for each pair of PWM output pins (P1A/P1C and P1B/P1D). The PWM output polarities must be selected before the PWM pin output drivers are enabled. Changing the polarity configuration while the PWM pin output drivers are enable is not recommended since it may result in damage to the application circuits.

The P1A, P1B, P1C and P1D output latches may not be in the proper states when the PWM module is initialized. Enabling the PWM pin output drivers at the same time as the Enhanced PWM modes may cause damage to the application circuit. The Enhanced PWM modes must be enabled in the proper Output mode and complete a full PWM cycle before enabling the PWM pin output drivers. The completion of a full PWM cycle is indicated by the TMR2IF bit of the PIR1 register being set as the second PWM period begins.

11.6.4 ENHANCED PWM AUTO-SHUTDOWN MODE

The PWM mode supports an Auto-Shutdown mode that will disable the PWM outputs when an external shutdown event occurs. Auto-Shutdown mode places the PWM output pins into a predetermined state. This mode is used to help prevent the PWM from damaging the application.

The auto-shutdown sources are selected using the ECCPAS<2:0> bits of the ECCPAS register. A shutdown event may be generated by:

- A logic '0' on the INT pin
- Comparator C1
- Comparator C2
- Setting the ECCPASE bit in firmware

A shutdown condition is indicated by the ECCPASE (Auto-Shutdown Event Status) bit of the ECCPAS register. If the bit is a '0', the PWM pins are operating normally. If the bit is a '1', the PWM outputs are in the shutdown state.

When a shutdown event occurs, two things happen:

The ECCPASE bit is set to '1'. The ECCPASE will remain set until cleared in firmware or an auto-restart occurs (see Section 11.6.5 "Auto-Restart Mode").

The enabled PWM pins are asynchronously placed in their shutdown states. The PWM output pins are grouped into pairs [P1A/P1C] and [P1B/P1D]. The state of each pin pair is determined by the PSSAC and PSSBD bits of the ECCPAS register. Each pin pair may be placed into one of three states:

- Drive logic '1'
- Drive logic '0'
- Tri-state (high-impedance)

REGISTER 11-3: ECCPAS: ENHANCED CAPTURE/COMPARE/PWM AUTO-SHUTDOWN CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
ECCPASE	ECCPAS2	ECCPAS1	ECCPAS0	PSSAC1	PSSAC0	PSSBD1	PSSBD0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	ECCPASE: ECCP Auto-Shutdown Event Status bit
	 1 = A shutdown event has occurred; ECCP outputs are in shutdown state 0 = ECCP outputs are operating
bit 6-4	ECCPAS<2:0>: ECCP Auto-shutdown Source Select bits
	000 = Auto-Shutdown is disabled
	001 = Comparator C1 output change
	010 = Comparator C2 output change ⁽¹⁾
	011 = Either Comparator C1 or C2 change
	100 = VIL on INT pin
	101 = VIL on INT pin or Comparator C1 change
	110 = VIL on INT pin or Comparator C2 change
	111 = VIL on INT pin or Comparator C1 or Comparator C2 change
bit 3-2	PSSACn: Pins P1A and P1C Shutdown State Control bits
	00 = Drive pins P1A and P1C to '0'
	01 = Drive pins P1A and P1C to '1'
	1x = Pins P1A and P1C tri-state
bit 1-0	PSSBDn: Pins P1B and P1D Shutdown State Control bits
	00 = Drive pins P1B and P1D to '0'
	01 = Drive pins P1B and P1D to '1'
	1x = Pins P1B and P1D tri-state

Note 1: If C2SYNC is enabled, the shutdown will be delayed by Timer1.

- Note 1: The auto-shutdown condition is a level-based signal, not an edge-based signal. As long as the level is present, the auto-shutdown will persist.
 - Writing to the ECCPASE bit is disabled while an auto-shutdown condition persists.
 - 3: Once the auto-shutdown condition has been removed and the PWM restarted (either through firmware or auto-restart) the PWM signal will always restart at the beginning of the next PWM period.

FIGURE 11-14: PWM AUTO-SHUTDOWN WITH FIRMWARE RESTART (PRSEN = 0)



11.6.5 AUTO-RESTART MODE

The Enhanced PWM can be configured to automatically restart the PWM signal once the auto-shutdown condition has been removed. Auto-restart is enabled by setting the PRSEN bit in the PWM1CON register.

If auto-restart is enabled, the ECCPASE bit will remain set as long as the auto-shutdown condition is active. When the auto-shutdown condition is removed, the ECCPASE bit will be cleared via hardware and normal operation will resume.

FIGURE 11-15: PWM AUTO-SHUTDOWN WITH AUTO-RESTART ENABLED (PRSEN = 1)


11.6.6 PROGRAMMABLE DEAD-BAND DELAY MODE

In Half-Bridge applications where all power switches are modulated at the PWM frequency, the power switches normally require more time to turn off than to turn on. If both the upper and lower power switches are switched at the same time (one turned on, and the other turned off), both switches may be on for a short period of time until one switch completely turns off. During this brief interval, a very high current (*shoot-through current*) will flow through both power switches, shorting the bridge supply. To avoid this potentially destructive shoot-through current from flowing during switching, turning on either of the power switches is normally delayed to allow the other switch to completely turn off.

In Half-Bridge mode, a digitally programmable dead-band delay is available to avoid shoot-through current from destroying the bridge power switches. The delay occurs at the signal transition from the non-active state to the active state. See Figure 11-16 for illustration. The lower seven bits of the associated PWM1CON register (Register 11-4) sets the delay period in terms of microcontroller instruction cycles (TcY or 4 Tosc).



FIGURE 11-17: EXAMPLE OF HALF-BRIDGE APPLICATIONS



R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0					
PRSEN	PRSEN PDC6 PDC5 PDC			PDC3	PDC2	PDC1	PDC0					
bit 7					•	bit 0						
Legend:												
R = Readable	bit	W = Writable	bit	U = Unimplen	nented bit, read	d as '0'						
-n = Value at F	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown					
bit 7 PRSEN: PWM Restart Enable bit												
1 = Upon auto-shutdown, the ECCPASE bit clears automatically once the shutdown event goes away; the PWM restarts automatically												

REGISTER 11-4: PWM1CON: ENHANCED PWM CONTROL REGISTER

0 = Upon auto-shutdown, ECCPASE must be cleared in software to restart the PWM

PDC<6:0>: PWM Delay Count bits bit 6-0

PDCn = Number of Fosc/4 (4 * Tosc) cycles between the scheduled time when a PWM signal should transition active and the actual time it transitions active

Note 1: Bit resets to '0' with Two-Speed Start-up and LP, XT or HS selected as the Oscillator mode or Fail-Safe mode is enabled.

11.6.7 PULSE STEERING MODE

In Single Output mode, pulse steering allows any of the PWM pins to be the modulated signal. Additionally, the same PWM signal can be simultaneously available on multiple pins.

Once the Single Output mode is selected (CCP1M<3:2> = 11 and P1M<1:0> = 00 of the CCP1CON register), the user firmware can bring out the same PWM signal to one, two, three or four output pins by setting the appropriate STR<D:A> bits of the PSTRCON register, as shown in Table 11-5.

Note:	The associated TRIS bits must be set to
	output ('0') to enable the pin output driver
	in order to see the PWM signal on the pin.

While the PWM Steering mode is active, CCP1M<1:0> bits of the CCP1CON register select the PWM output polarity for the P1<D:A> pins.

The PWM auto-shutdown operation also applies to PWM Steering mode as described in **Section 11.6.4 "Enhanced PWM Auto-shutdown mode"**. An auto-shutdown event will only affect pins that have PWM outputs enabled.

REGISTER 11-5: PSTRCON: PULSE STEERING CONTROL REGISTER⁽¹⁾

U-0	U-0	U-0	R/W-0	R/W-0	R/W-0 R/W-0		R/W-1
_	—	—	STRSYNC	STRD	STRC	STRB	STRA
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-5	Unimplemented: Read as '0'
bit 4	STRSYNC: Steering Sync bit
	 1 = Output steering update occurs on next PWM period 0 = Output steering update occurs at the beginning of the instruction cycle boundary
bit 3	STRD: Steering Enable bit D
	1 = P1D pin has the PWM waveform with polarity control from CCPxM<1:0>
	0 = P1D pin is assigned to port pin
bit 2	STRC: Steering Enable bit C
	1 = P1C pin has the PWM waveform with polarity control from CCPxM<1:0>
	0 = P1C pin is assigned to port pin
bit 1	STRB: Steering Enable bit B
	1 = P1B pin has the PWM waveform with polarity control from CCPxM<1:0>
	0 = P1B pin is assigned to port pin
bit 0	STRA: Steering Enable bit A
	1 = P1A pin has the PWM waveform with polarity control from CCPxM<1:0>
	0 = P1A pin is assigned to port pin
Noto 1	The DW/M Steering mode is available only when the CCD1CON register bits CCD1M c^{2} : $2 - 11$

Note 1: The PWM Steering mode is available only when the CCP1CON register bits CCP1M<3:2> = 11 and P1M<1:0> = 00.



FIGURE 11-18: SIMPLIFIED STEERING BLOCK DIAGRAM

2: Single PWM output requires setting at least one of the STRx bits.

11.6.7.1 Steering Synchronization

The STRSYNC bit of the PSTRCON register gives the user two selections of when the steering event will happen. When the STRSYNC bit is '0', the steering event will happen at the end of the instruction that writes to the PSTRCON register. In this case, the output signal at the P1<D:A> pins may be an incomplete PWM waveform. This operation is useful when the user firmware needs to immediately remove a PWM signal from the pin.

When the STRSYNC bit is '1', the effective steering update will happen at the beginning of the next PWM period. In this case, steering on/off the PWM output will always produce a complete PWM waveform. Figures 11-19 and 11-20 illustrate the timing diagrams of the PWM steering depending on the STRSYNC setting.

FIGURE 11-19: EXAMPLE OF STEERING EVENT AT END OF INSTRUCTION (STRSYNC = 0)



FIGURE 11-20: EXAMPLE OF STEERING EVENT AT BEGINNING OF INSTRUCTION (STRSYNC = 1)



Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
CCP1CON	P1M1	P1M0	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	0000 0000	0000 0000
CCP2CON	—	_	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	00 0000	00 0000
CCPR1L	Capture/Cor	mpare/PWM F	Register 1 Lov	v Byte (LSB)					XXXX XXXX	XXXX XXXX
CCPR1H	Capture/Cor	mpare/PWM F	Register 1 Hig	h Byte (MSB))				XXXX XXXX	XXXX XXXX
CCPR2L	Capture/Cor	mpare/PWM F	Register 2 Lov	v Byte (LSB)					XXXX XXXX	XXXX XXXX
CCPR2H	Capture/Cor	mpare/PWM F	Register 2 Hig	h Byte (MSB))				XXXX XXXX	XXXX XXXX
CM2CON1	MC1OUT	MC2OUT	C1RSEL	C2RSEL	—	—	T1GSS	C2SYNC	000010	000010
INTCON	GIE	PEIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF	x000 0000	0000 000x
PIE1	_	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	-000 0000	-000 0000
PIE2	OSFIE	C2IE	C1IE	EEIE	BCLIE	ULPWUIE	_	CCP2IE	0000 00-0	0000 00-0
PIR1	_	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	-000 0000	-000 0000
PIR2	OSFIF	C2IF	C1IF	EEIF	BCLIF	ULPWUIF	_	CCP2IF	0000 00-0	0000 00-0
T1CON	T1GINV	TMR1GE	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N	0000 0000	0000 0000
TMR1L	Holding Reg	jister for the L	east Significa.	int Byte of the	16-bit TMR1	Register			XXXX XXXX	xxxx xxxx
TMR1H	Holding Reg	gister for the N	lost Significa	nt Byte of the	16-bit TMR1	Register			xxxx xxxx	xxxx xxxx
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111

TABLE 11-6: REGISTERS ASSOCIATED WITH CAPTURE, COMPARE AND TIMER1

Legend: -= Unimplemented locations, read as '0', u = unchanged, x = unknown. Shaded cells are not used by the Capture and Compare.

TABLE 11-7: REGISTERS ASSOCIATED WITH PWM AND TIMER2

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
CCP1CON	P1M1	P1M0	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	0000 0000	0000 0000
CCP2CON	—		DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	00 0000	00 0000
ECCPAS	ECCPASE	ECCPAS2	ECCPAS1	ECCPAS0	PSSAC1	PSSAC0	PSSBD1	PSSBD0	0000 0000	0000 0000
INTCON	GIE	PEIE	TOIE	INTE	RBIE	T0IF	INTF	RBIF	x000 0000	0000 000x
PR2	Timer2 Perio	od Register							1111 1111	1111 1111
PSTRCON	—	_	_	STRSYNC	STRD	STRC	STRB	STRA	0 0001	0 0001
PWM1CON	PRSEN	PDC6	PDC5	PDC4	PDC3	PDC2	PDC1	PDC0	0000 0000	0000 0000
T2CON	_	TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0	-000 0000	-000 0000
TMR2	Timer2 Mod	ule Register							0000 0000	0000 0000
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	1111 1111	1111 1111
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111
TRISD	TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0	1111 1111	1111 1111

Legend: -= Unimplemented locations, read as '0', u = unchanged, x = unknown. Shaded cells are not used by the PWM.

12.0 ENHANCED UNIVERSAL SYNCHRONOUS ASYNCHRONOUS RECEIVER TRANSMITTER (EUSART)

The Enhanced Universal Synchronous Asynchronous Receiver Transmitter (EUSART) module is a serial I/O communications peripheral. It contains all the clock generators, shift registers and data buffers necessary to perform an input or output serial data transfer independent of device program execution. The EUSART, also known as a Serial Communications Interface (SCI), can be configured as a full-duplex asynchronous system or half-duplex synchronous Full-Duplex mode system. is useful for communications with peripheral systems, such as CRT terminals and personal computers. Half-Duplex Synchronous mode is intended for communications with peripheral devices, such as A/D or D/A integrated circuits, serial EEPROMs or other microcontrollers. These devices typically do not have internal clocks for baud rate generation and require the external clock signal provided by a master synchronous device.

The EUSART module includes the following capabilities:

- Full-duplex asynchronous transmit and receive
- Two-character input buffer
- One-character output buffer
- Programmable 8-bit or 9-bit character length
- Address detection in 9-bit mode
- Input buffer overrun error detection
- Received character framing error detection
- Half-duplex synchronous master
- Half-duplex synchronous slave
- Programmable clock polarity in synchronous modes

The EUSART module implements the following additional features, making it ideally suited for use in Local Interconnect Network (LIN) bus systems:

- · Automatic detection and calibration of the baud rate
- Wake-up on Break reception
- 13-bit Break character transmit

Block diagrams of the EUSART transmitter and receiver are shown in Figure 12-1 and Figure 12-2.

FIGURE 12-1: EUSART TRANSMIT BLOCK DIAGRAM



FIGURE 12-2: EUSART RECEIVE BLOCK DIAGRAM



The operation of the EUSART module is controlled through three registers:

- Transmit Status and Control (TXSTA)
- Receive Status and Control (RCSTA)
- Baud Rate Control (BAUDCTL)

These registers are detailed in Register 12-1, Register 12-2 and Register 12-3, respectively.

12.1 EUSART Asynchronous Mode

The EUSART transmits and receives data using the standard non-return-to-zero (NRZ) format. NRZ is implemented with two levels: a VOH mark state which represents a '1' data bit, and a VOL space state which represents a '0' data bit. NRZ refers to the fact that consecutively transmitted data bits of the same value stay at the output level of that bit without returning to a neutral level between each bit transmission. An NRZ transmission port idles in the mark state. Each character transmission consists of one Start bit followed by eight or nine data bits and is always terminated by one or more Stop bits. The Start bit is always a space and the Stop bits are always marks. The most common data format is 8 bits. Each transmitted bit persists for a period of 1/(Baud Rate). An on-chip dedicated 8-bit/16-bit Baud Rate Generator is used to derive standard baud rate frequencies from the system oscillator. See Table 12-5 for examples of baud rate configurations.

The EUSART transmits and receives the LSb first. The EUSART's transmitter and receiver are functionally independent, but share the same data format and baud rate. Parity is not supported by the hardware, but can be implemented in software and stored as the ninth data bit.

12.1.1 EUSART ASYNCHRONOUS TRANSMITTER

The EUSART transmitter block diagram is shown in Figure 12-1. The heart of the transmitter is the serial Transmit Shift Register (TSR), which is not directly accessible by software. The TSR obtains its data from the transmit buffer, which is the TXREG register.

12.1.1.1 Enabling the Transmitter

The EUSART transmitter is enabled for asynchronous operations by configuring the following three control bits:

- TXEN = 1
- SYNC = 0
- SPEN = 1

All other EUSART control bits are assumed to be in their default state.

Setting the TXEN bit of the TXSTA register enables the transmitter circuitry of the EUSART. Clearing the SYNC bit of the TXSTA register configures the EUSART for asynchronous operation. Setting the SPEN bit of the RCSTA register enables the EUSART and automatically configures the TX/CK I/O pin as an output. If the TX/CK pin is shared with an analog peripheral the analog I/O function must be disabled by clearing the corresponding ANSEL bit.

- Note 1: When the SPEN bit is set the RX/DT I/O pin is automatically configured as an input, regardless of the state of the corresponding TRIS bit and whether or not the EUSART receiver is enabled. The RX/DT pin data can be read via a normal PORT read but PORT latch data output is precluded.
 - **2:** The TXIF transmitter interrupt flag is set when the TXEN enable bit is set.

12.1.1.2 Transmitting Data

A transmission is initiated by writing a character to the TXREG register. If this is the first character, or the previous character has been completely flushed from the TSR, the data in the TXREG is immediately transferred to the TSR register. If the TSR still contains all or part of a previous character, the new character data is held in the TXREG until the Stop bit of the previous character has been transmitted. The pending character in the TXREG is then transferred to the TSR in one TCY immediately following the Stop bit sequence commences immediately following the transfer of the data to the TSR from the TXREG.

12.1.1.3 Transmit Interrupt Flag

The TXIF interrupt flag bit of the PIR1 register is set whenever the EUSART transmitter is enabled and no character is being held for transmission in the TXREG. In other words, the TXIF bit is only clear when the TSR is busy with a character and a new character has been queued for transmission in the TXREG. The TXIF flag bit is not cleared immediately upon writing TXREG. TXIF becomes valid in the second instruction cycle following the write execution. Polling TXIF immediately following the TXREG write will return invalid results. The TXIF bit is read-only, it cannot be set or cleared by software.

The TXIF interrupt can be enabled by setting the TXIE interrupt enable bit of the PIE1 register. However, the TXIF flag bit will be set whenever the TXREG is empty, regardless of the state of TXIE enable bit.

To use interrupts when transmitting data, set the TXIE bit only when there is more data to send. Clear the TXIE interrupt enable bit upon writing the last character of the transmission to the TXREG.

12.1.1.4 TSR Status

The TRMT bit of the TXSTA register indicates the status of the TSR register. This is a read-only bit. The TRMT bit is set when the TSR register is empty and is cleared when a character is transferred to the TSR register from the TXREG. The TRMT bit remains clear until all bits have been shifted out of the TSR register. No interrupt logic is tied to this bit, so the user has to poll this bit to determine the TSR status.

Note:	The TSR register is not mapped in data
	memory, so it is not available to the user.

12.1.1.5 Transmitting 9-Bit Characters

The EUSART supports 9-bit character transmissions. When the TX9 bit of the TXSTA register is set the EUSART will shift 9 bits out for each character transmitted. The TX9D bit of the TXSTA register is the ninth, and Most Significant, data bit. When transmitting 9-bit data, the TX9D data bit must be written before writing the 8 Least Significant bits into the TXREG. All nine bits of data will be transferred to the TSR shift register immediately after the TXREG is written.

A special 9-bit Address mode is available for use with multiple receivers. See **Section 12.1.2.7** "Address **Detection**" for more information on the Address mode.

12.1.1.6 Asynchronous Transmission Set-up:

- 1. Initialize the SPBRGH, SPBRG register pair and the BRGH and BRG16 bits to achieve the desired baud rate (see Section 12.3 "EUSART Baud Rate Generator (BRG)").
- 2. Enable the asynchronous serial port by clearing the SYNC bit and setting the SPEN bit.
- 3. If 9-bit transmission is desired, set the TX9 control bit. A set ninth data bit will indicate that the 8 Least Significant data bits are an address when the receiver is set for address detection.
- 4. Enable the transmission by setting the TXEN control bit. This will cause the TXIF interrupt bit to be set.
- If interrupts are desired, set the TXIE interrupt enable bit. An interrupt will occur immediately provided that the GIE and PEIE bits of the INTCON register are also set.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded into the TX9D data bit.
- 7. Load 8-bit data into the TXREG register. This will start the transmission.



FIGURE 12-4: ASYNCHRONOUS TRANSMISSION (BACK-TO-BACK)



FIGURE 12-3: ASYNCHRONOUS TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
BAUDCTL	ABDOVF	RCIDL	_	SCKP	BRG16	_	WUE	ABDEN	01-0 0-00	01-0 0-00
INTCON	GIE	PEIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	0000 000x
PIE1	_	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	-000 0000	-000 0000
PIR1	_	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	-000 0000	-000 0000
RCREG	EUSART I	Receive Da	ita Register						0000 0000	0000 0000
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000x	0000 000x
SPBRG	BRG7	BRG6	BRG5	BRG4	BRG3	BRG2	BRG1	BRG0	0000 0000	0000 0000
SPBRGH	BRG15	BRG14	BRG13	BRG12	BRG11	BRG10	BRG9	BRG8	0000 0000	0000 0000
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111
TXREG	EUSART	Transmit Da	ata Registe	r					0000 0000	0000 0000
TXSTA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	0000 0010	0000 0010

TABLE 12-1: REGISTERS ASSOCIATED WITH ASYNCHRONOUS TRANSMISSION

Legend: x = unknown, - = unimplemented read as '0'. Shaded cells are not used for Asynchronous Transmission.

12.1.2 EUSART ASYNCHRONOUS RECEIVER

The Asynchronous mode would typically be used in RS-232 systems. The receiver block diagram is shown in Figure 12-2. The data is received on the RX/DT pin and drives the data recovery block. The data recovery block is actually a high-speed shifter operating at 16 times the baud rate, whereas the serial Receive Shift Register (RSR) operates at the bit rate. When all 8 or 9 bits of the character have been shifted in, they are immediately transferred to a two character First-In-First-Out (FIFO) memory. The FIFO buffering allows reception of two complete characters and the start of a third character before software must start servicing the EUSART receiver. The FIFO and RSR registers are not directly accessible by software. Access to the received data is via the RCREG register.

12.1.2.1 Enabling the Receiver

The EUSART receiver is enabled for asynchronous operation by configuring the following three control bits:

- CREN = 1
- SYNC = 0
- SPEN = 1

All other EUSART control bits are assumed to be in their default state.

Setting the CREN bit of the RCSTA register enables the receiver circuitry of the EUSART. Clearing the SYNC bit of the TXSTA register configures the EUSART for asynchronous operation. Setting the SPEN bit of the RCSTA register enables the EUSART and automatically configures the RX/DT I/O pin as an input. If the RX/DT pin is shared with an analog peripheral the analog I/O function must be disabled by clearing the corresponding ANSEL bit.

Note: When the SPEN bit is set the TX/CK I/O pin is automatically configured as an output, regardless of the state of the corresponding TRIS bit and whether or not the EUSART transmitter is enabled. The PORT latch is disconnected from the output driver so it is not possible to use the TX/CK pin as a general purpose output.

12.1.2.2 Receiving Data

The receiver data recovery circuit initiates character reception on the falling edge of the first bit. The first bit, also known as the Start bit, is always a zero. The data recovery circuit counts one-half bit time to the center of the Start bit and verifies that the bit is still a zero. If it is not a zero then the data recovery circuit aborts character reception, without generating an error, and resumes looking for the falling edge of the Start bit. If the Start bit zero verification succeeds then the data recovery circuit counts a full bit time to the center of the next bit. The bit is then sampled by a majority detect circuit and the resulting '0' or '1' is shifted into the RSR. This repeats until all data bits have been sampled and shifted into the RSR. One final bit time is measured and the level sampled. This is the Stop bit, which is always a '1'. If the data recovery circuit samples a '0' in the Stop bit position then a framing error is set for this character, otherwise the framing error is cleared for this character. See Section 12.1.2.4 "Receive Framing Error" for more information on framing errors.

Immediately after all data bits and the Stop bit have been received, the character in the RSR is transferred to the EUSART receive FIFO and the RCIF interrupt flag bit of the PIR1 register is set. The top character in the FIFO is transferred out of the FIFO by reading the RCREG register.

Note:	If the receive FIFO is overrun, no additional characters will be received until the overrun condition is cleared. See Section 12.1.2.	ן ז ז
	"Receive Overrun Error" for more	ì

12.1.2.3 Receive Interrupts

The RCIF interrupt flag bit of the PIR1 register is set whenever the EUSART receiver is enabled and there is an unread character in the receive FIFO. The RCIF interrupt flag bit is read-only, it cannot be set or cleared by software.

RCIF interrupts are enabled by setting the following bits:

- RCIE interrupt enable bit of the PIE1 register
- PEIE peripheral interrupt enable bit of the INTCON register
- GIE global interrupt enable bit of the INTCON register

The RCIF interrupt flag bit will be set when there is an unread character in the FIFO, regardless of the state of interrupt enable bits.

12.1.2.4 Receive Framing Error

Each character in the receive FIFO buffer has a corresponding framing error Status bit. A framing error indicates that a Stop bit was not seen at the expected time. The framing error status is accessed via the FERR bit of the RCSTA register. The FERR bit represents the status of the top unread character in the receive FIFO. Therefore, the FERR bit must be read before reading the RCREG.

The FERR bit is read-only and only applies to the top unread character in the receive FIFO. A framing error (FERR = 1) does not preclude reception of additional characters. It is not necessary to clear the FERR bit. Reading the next character from the FIFO buffer will advance the FIFO to the next character and the next corresponding framing error.

The FERR bit can be forced clear by clearing the SPEN bit of the RCSTA register which resets the EUSART. Clearing the CREN bit of the RCSTA register does not affect the FERR bit. A framing error by itself does not generate an interrupt.

Note:	If all receive characters in the receive
	FIFO have framing errors, repeated reads
	of the RCREG will not clear the FERR bit.

12.1.2.5 Receive Overrun Error

The receive FIFO buffer can hold two characters. An overrun error will be generated If a third character, in its entirety, is received before the FIFO is accessed. When this happens the OERR bit of the RCSTA register is set. The characters already in the FIFO buffer can be read but no additional characters will be received until the error is cleared. The error must be cleared by either clearing the CREN bit of the RCSTA register or by resetting the EUSART by clearing the SPEN bit of the RCSTA register.

12.1.2.6 Receiving 9-bit Characters

The EUSART supports 9-bit character reception. When the RX9 bit of the RCSTA register is set the EUSART will shift 9 bits into the RSR for each character received. The RX9D bit of the RCSTA register is the ninth and Most Significant data bit of the top unread character in the receive FIFO. When reading 9-bit data from the receive FIFO buffer, the RX9D data bit must be read before reading the 8 Least Significant bits from the RCREG.

12.1.2.7 Address Detection

A special Address Detection mode is available for use when multiple receivers share the same transmission line, such as in RS-485 systems. Address detection is enabled by setting the ADDEN bit of the RCSTA register.

Address detection requires 9-bit character reception. When address detection is enabled, only characters with the ninth data bit set will be transferred to the receive FIFO buffer, thereby setting the RCIF interrupt bit. All other characters will be ignored.

Upon receiving an address character, user software determines if the address matches its own. Upon address match, user software must disable address detection by clearing the ADDEN bit before the next Stop bit occurs. When user software detects the end of the message, determined by the message protocol used, software places the receiver back into the Address Detection mode by setting the ADDEN bit.

12.1.2.8 Asynchronous Reception Set-up:

- Initialize the SPBRGH, SPBRG register pair and the BRGH and BRG16 bits to achieve the desired baud rate (see Section 12.3 "EUSART Baud Rate Generator (BRG)").
- 2. Enable the serial port by setting the SPEN bit. The SYNC bit must be clear for asynchronous operation.
- 3. If interrupts are desired, set the RCIE interrupt enable bit and set the GIE and PEIE bits of the INTCON register.
- 4. If 9-bit reception is desired, set the RX9 bit.
- 5. Enable reception by setting the CREN bit.
- 6. The RCIF interrupt flag bit will be set when a character is transferred from the RSR to the receive buffer. An interrupt will be generated if the RCIE interrupt enable bit was also set.
- 7. Read the RCSTA register to get the error flags and, if 9-bit data reception is enabled, the ninth data bit.
- 8. Get the received 8 Least Significant data bits from the receive buffer by reading the RCREG register.
- 9. If an overrun occurred, clear the OERR flag by clearing the CREN receiver enable bit.

12.1.2.9 9-bit Address Detection Mode Set-up

This mode would typically be used in RS-485 systems. To set up an Asynchronous Reception with Address Detect Enable:

- Initialize the SPBRGH, SPBRG register pair and the BRGH and BRG16 bits to achieve the desired baud rate (see Section 12.3 "EUSART Baud Rate Generator (BRG)").
- 2. Enable the serial port by setting the SPEN bit. The SYNC bit must be clear for asynchronous operation.
- 3. If interrupts are desired, set the RCIE interrupt enable bit and set the GIE and PEIE bits of the INTCON register.
- 4. Enable 9-bit reception by setting the RX9 bit.
- 5. Enable address detection by setting the ADDEN bit.
- 6. Enable reception by setting the CREN bit.
- 7. The RCIF interrupt flag bit will be set when a character with the ninth bit set is transferred from the RSR to the receive buffer. An interrupt will be generated if the RCIE interrupt enable bit was also set.
- 8. Read the RCSTA register to get the error flags. The ninth data bit will always be set.
- Get the received 8 Least Significant data bits from the receive buffer by reading the RCREG register. Software determines if this is the device's address.
- 10. If an overrun occurred, clear the OERR flag by clearing the CREN receiver enable bit.
- 11. If the device has been addressed, clear the ADDEN bit to allow all received data into the receive buffer and generate interrupts.



FIGURE 12-5: ASYNCHRONOUS RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
BAUDCTL	ABDOVF	RCIDL	—	SCKP	BRG16	—	WUE	ABDEN	01-0 0-00	01-0 0-00
INTCON	GIE	PEIE	TOIE	INTE	RBIE	T0IF	INTF	RBIF	x000 000x	x000 0000
PIE1	_	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	-000 0000	-000 0000
PIR1	_	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	-000 0000	-000 0000
RCREG	EUSART	Receive Da	ta Register						0000 0000	0000 0000
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	x000 000x	x000 0000
SPBRG	BRG7	BRG6	BRG5	BRG4	BRG3	BRG2	BRG1	BRG0	0000 0000	0000 0000
SPBRGH	BRG15	BRG14	BRG13	BRG12	BRG11	BRG10	BRG9	BRG8	0000 0000	0000 0000
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111
TXREG	EUSART Transmit Data Register							0000 0000	0000 0000	
TXSTA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	0000 0010	0000 0010
Lonondi			malamanta		¹ Chodod a		unand for A	over ob ron ou	Decention	

REGISTERS ASSOCIATED WITH ASYNCHRONOUS RECEPTION TABLE 12-2:

Legend: x = unknown, - = unimplemented read as '0'. Shaded cells are not used for Asynchronous Reception.

12.2 Clock Accuracy with Asynchronous Operation

The factory calibrates the internal oscillator block output (INTOSC). However, the INTOSC frequency may drift as VDD or temperature changes, and this directly affects the asynchronous baud rate. Two methods may be used to adjust the baud rate clock, but both require a reference clock source of some kind. The first (preferred) method uses the OSCTUNE register to adjust the INTOSC output. Adjusting the value in the OSCTUNE register allows for fine resolution changes to the system clock source. See **4.5** "Internal Clock Modes" for more information.

The other method adjusts the value in the Baud Rate Generator. This can be done automatically with the Auto-Baud Detect feature (see **Section 12.3.1 "Auto-Baud Detect**"). There may not be fine enough resolution when adjusting the Baud Rate Generator to compensate for a gradual change in the peripheral clock frequency.

REGISTER 12-1: TXSTA: TRANSMIT STATUS AND CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-1	R/W-0
CSRC	TX9	TXEN ⁽¹⁾	SYNC	SENDB	BRGH	TRMT	TX9D
bit 7							bit 0
Legend:							
R = Readable b	bit	W = Writable bi	t	U = Unimplem	ented bit, read as	ʻ0'	
-n = Value at PO	OR	'1' = Bit is set		'0' = Bit is clea	red	x = Bit is unkno	wn
bit 7	CSRC: Clock	Source Select bit					
	Asynchronous	mode:					
	Don't care						
	Synchronous r	<u>node</u> :					
	1 = Master m	node (clock gener	ated internally	from BRG)			
	0 = Slave models	ode (clock from ex	(ternal source)				
bit 6	TX9: 9-bit Trar	nsmit Enable bit					
	1 = Selects S	<i>bit transmission</i>					
hit F							
DIUD	1 - Transmit						
	0 = Transmit	disabled					
bit 4	SYNC: EUSAR	RT Mode Select b	it				
	1 = Synchron	ious mode					
	0 = Asynchro	nous mode					
bit 3	SENDB: Send	Break Character	bit				
	<u>Asynchronous</u>	mode:					
	1 = Send Syr	nc Break on next t	ransmission (o	leared by hardwa	are upon complet	on)	
	0 = Sync Bre	ak transmission c	ompleted				
	Don't care	<u>noue</u> .					
hit 2	BBCH: High B	aud Pata Salact I	oit				
DIT Z	Asynchronous	mode:	JI				
	1 = High spee	ed					
	0 = Low spee	ed					
	Synchronous r	<u>mode:</u>					
	Unused in this	mode					
bit 1	TRMT: Transm	nit Shift Register S	Status bit				
	1 = TSR emp	oty					
	0 = TSR full						
bit 0	TX9D: Ninth b	it of Transmit Data	a 				
	Can be addres	ss/data bit or a pai	rity bit.				

Note 1: SREN/CREN overrides TXEN in Sync mode.

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-0	R-0	R-x
SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D
bit 7							bit 0
Legend:							
R = Readable	bit	W = Writable	bit	U = Unimpler	mented bit, read	as '0'	
-n = Value at F	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkn	nown
bit 7	SPEN: Serial 1 = Serial po	Port Enable bi rt enabled (cor	t ıfigures RX/D	T and TX/CK p	ins as serial po	rt pins)	
	0 = Serial po	rt disabled (hel	d in Reset)				
bit 6	RX9: 9-bit Re	ceive Enable b	bit				
	1 = Selects 9 0 = Selects 8	B-bit reception B-bit reception					
bit 5	SREN: Single	e Receive Enat	ole bit				
	Asynchronou:	<u>s mode</u> :					
	Don't care						
	Synchronous	<u>mode – Maste</u>	<u>r</u> :				
	1 = Enables 0 = Disables	single receive					
	This bit is clea	ared after rece	otion is compl	lete.			
	Synchronous	mode – Slave					
	Don't care						
bit 4	CREN: Contin	nuous Receive	Enable bit				
	<u>Asynchronous</u>	<u>s mode</u> . receiver					
	0 = Disables	receiver					
	Synchronous	mode:					
	1 = Enables 0 = Disables	continuous rec continuous rec	eive until enal ceive	ble bit CREN is	s cleared (CREN	I overrides SRI	EN)
bit 3	ADDEN: Add	ress Detect En	able bit				
	Asynchronous	<u>s mode 9-bit (F</u>	<u> X9 = 1)</u> :				
	1 = Enables 0 = Disables <u>Asynchronous</u>	address detect address detec <u>s mode 8-bit (</u> R	ion, enable in tion, all bytes <u>XX9 = 0)</u> :	terrupt and loa are received a	d the receive bund ninth bit can	Iffer when RSR be used as pai	l<8> is set rity bit
	Don't care						
bit 2	FERR: Frami	ng Error bit					
	1 = Framing 0 = No framin	error (can be u ng error	pdated by rea	ading RCREG	register and rec	eive next valid	byte)
bit 1	OERR: Overr	un Error bit					
	1 = Overrun 0 = No overr	error (can be c un error	leared by clea	aring bit CREN)		
bit 0	RX9D: Ninth	bit of Received	Data				
	This can be a	ddress/data bit	or a parity bi	t and must be o	calculated by us	er firmware.	

REGISTER 12-2: RCSTA: RECEIVE STATUS AND CONTROL REGISTER⁽¹⁾

R-0	R-1	U-0	R/W-0	R/W-0	U-0	R/W-0	R/W-0
ABDOVF	RCIDL	_	SCKP	BRG16		WUE	ABDEN
bit 7				•			bit 0
Legend:							
R = Readable	bit	W = Writable I	oit	U = Unimple	mented bit, read	as '0'	
-n = Value at F	POR	'1' = Bit is set		'0' = Bit is cle	eared	x = Bit is unkr	nown
bit 7	ABDOVF: Au	to-Baud Detect	Overflow bit				
	Asynchronous	<u>s mode</u> :	a d				
	$\perp = Auto-bauco0 = Auto-bauco$	d timer overnow	vea				
	Synchronous	mode:					
	Don't care						
bit 6	RCIDL: Rece	ive Idle Flag bit					
	Asynchronous	<u>s mode</u> :					
	1 = Receiver 0 = Start bit h	as been receiv	ed and the re	ceiver is receiv	vina		
	Synchronous	mode:					
	Don't care						
bit 5	Unimplemen	ted: Read as 'o)'				
bit 4	SCKP: Synch	ronous Clock F	olarity Selec	t bit			
	Asynchronous	<u>s mode</u> :					
	1 = Transmit i 0 = Transmit i	inverted data to non-inverted da	the RB7/TX/ ta to the RB7	CK pin //TX/CK pin			
	Synchronous	mode:					
	1 = Data is closed	ocked on rising	edge of the c	clock			
bit 3	BRG16: 16-bi	it Baud Rate G	enerator bit	CIOCIC			
	1 = 16-bit Ba	ud Rate Gener	ator is used				
	0 = 8-bit Bau	d Rate Genera	tor is used				
bit 2	Unimplemen	ted: Read as 'o)'				
bit 1	WUE: Wake-u	up Enable bit					
	Asynchronous	<u>s mode</u> :					
	1 = Receiver i	is waiting for a f	alling edge. N	lo character w	ill be received by	te RCIF will be	e set. WUE will
	automatic	cally clear after	RCIF IS Set.				
	Synchronous	mode:	intany				
	Don't care						
bit 0	ABDEN: Auto	-Baud Detect E	nable bit				
	Asynchronous	<u>s mode</u> :					
	1 = Auto-Bau	Id Detect mode	is enabled (o	clears when au	ito-baud is comp	olete)	
	0 = Auto-Bau Synchronous	Id Detect mode	is disabled				
	Don't care						

REGISTER 12-3: BAUDCTL: BAUD RATE CONTROL REGISTER

12.3 EUSART Baud Rate Generator (BRG)

The Baud Rate Generator (BRG) is an 8-bit or 16-bit timer that is dedicated to the support of both the asynchronous and synchronous EUSART operation. By default, the BRG operates in 8-bit mode. Setting the BRG16 bit of the BAUDCTL register selects 16-bit mode.

The SPBRGH, SPBRG register pair determines the period of the free running baud rate timer. In Asynchronous mode the multiplier of the baud rate period is determined by both the BRGH bit of the TXSTA register and the BRG16 bit of the BAUDCTL register. In Synchronous mode, the BRGH bit is ignored.

Table 12-3 contains the formulas for determining the baud rate. Example 12-1 provides a sample calculation for determining the baud rate and baud rate error.

Typical baud rates and error values for various asynchronous modes have been computed for your convenience and are shown in Table 12-3. It may be advantageous to use the high baud rate (BRGH = 1), or the 16-bit BRG (BRG16 = 1) to reduce the baud rate error. The 16-bit BRG mode is used to achieve slow baud rates for fast oscillator frequencies.

Writing a new value to the SPBRGH, SPBRG register pair causes the BRG timer to be reset (or cleared). This ensures that the BRG does not wait for a timer overflow before outputting the new baud rate. If the system clock is changed during an active receive operation, a receive error or data loss may result. To avoid this problem, check the status of the RCIDL bit to make sure that the receive operation is Idle before changing the system clock.

EXAMPLE 12-1: CALCULATING BAUD RATE ERROR

For a device with Fosc of 16 MHz, desired baud rate of 9600, Asynchronous mode, 8-bit BRG:
Desired Baud Rate = $\frac{FOSC}{64([SPBRGH:SPBRG] + 1)}$
Solving for SPBRGH:SPBRG:
$X = \frac{\frac{Fosc}{Desired Baud Rate}}{64} - 1$
$=\frac{\frac{16000000}{9600}}{64}-1$
= [25.042] = 25
Calculated Baud Rate = $\frac{16000000}{64(25+1)}$
= 9615
Error = Calc. Baud Rate – Desired Baud Rate Desired Baud Rate
$= \frac{(9615 - 9600)}{9600} = 0.16\%$

	Configuration Bits			Baud Pate Formula				
SYNC	BRG16	BRGH	BRG/EUSART Mode	Bauu Kale Formula				
0	0	0	8-bit/Asynchronous	Fosc/[64 (n+1)]				
0	0	1	8-bit/Asynchronous					
0	1	0	16-bit/Asynchronous	FOSC/[16 (n+1)]				
0	1	1	16-bit/Asynchronous					
1	0	x	8-bit/Synchronous	Fosc/[4 (n+1)]				
1	1	x	16-bit/Synchronous					

TABLE 12-3: BAUD RATE FORMULAS

Legend: x = don't care, n = value of SPBRGH, SPBRG register pair

TABLE 12-4: REGISTERS ASSOCIATED WITH THE BAUD RATE GENERATOR

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
BAUDCTL	ABDOVF	RCIDL	—	SCKP	BRG16		WUE	ABDEN	01-0 0-00	01-0 0-00
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	x000 0000	x000 000x
SPBRG	BRG7	BRG6	BRG5	BRG4	BRG3	BRG2	BRG1	BRG0	0000 0000	0000 0000
SPBRGH	BRG15	BRG14	BRG13	BRG12	BRG11	BRG10	BRG9	BRG8	0000 0000	0000 0000
TXSTA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	0000 0010	0000 0010

Legend: x = unknown, - = unimplemented read as '0'. Shaded cells are not used for the Baud Rate Generator.

					SYNC	C = 0, BRGI	i = 0, BRC	G16 = 0				
BAUD	Fost	c = 20.00	0 MHz	Foso	: = 18.43	2 MHz	Fosc	= 11.059	92 MHz	Fosc = 8.000 MHz		
RATE 300	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)
300	_	_	_	_	_			_	_		_	
1200	1221	1.73	255	1200	0.00	239	1200	0.00	143	1202	0.16	103
2400	2404	0.16	129	2400	0.00	119	2400	0.00	71	2404	0.16	51
9600	9470	-1.36	32	9600	0.00	29	9600	0.00	17	9615	0.16	12
10417	10417	0.00	29	10286	-1.26	27	10165	-2.42	16	10417	0.00	11
19.2k	19.53k	1.73	15	19.20k	0.00	14	19.20k	0.00	8	—	_	_
57.6k	_	—	_	57.60k	0.00	7	57.60k	0.00	2	—	—	—
115.2k	—		_	—	_	_	—	_	_	—	_	_

TABLE 12-5: BAUD RATES FOR ASYNCHRONOUS MODES

		SYNC = 0, BRGH = 0, BRG16 = 0												
BAUD	Fos	c = 4.00	0 MHz	Fosc	: = 3.686	4 MHz	Fos	c = 2.00	0 MHz	Fos	c = 1.00	0 MHz		
RATE	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)		
300	300	0.16	207	300	0.00	191	300	0.16	103	300	0.16	51		
1200	1202	0.16	51	1200	0.00	47	1202	0.16	25	1202	0.16	12		
2400	2404	0.16	25	2400	0.00	23	2404	0.16	12	—	—			
9600	—	—		9600	0.00	5	—	—	—	—	—			
10417	10417	0.00	5	—	_	_	10417	0.00	2	—	_	_		
19.2k	—	_	_	19.20k	0.00	2	—	—	—	—	_	_		
57.6k	—	—	_	57.60k	0.00	0	—	—	—	—	—	_		
115.2k	—	_	_	—	—	_	_	—	_	—	_	_		

	SYNC = 0, BRGH = 1, BRG16 = 0											
BAUD	Fosc	= 20.00	0 MHz	Fosc = 18.432 MHz			Fosc = 11.0592 MHz			Fosc = 8.000 MHz		
RATE	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)
300	—		_	—	_	_	_	_	—		_	—
1200	—	—	—	—	—	—	—	—	—	—	—	—
2400	—	_	—	—	_	—	—	_	_	2404	0.16	207
9600	9615	0.16	129	9600	0.00	119	9600	0.00	71	9615	0.16	51
10417	10417	0.00	119	10378	-0.37	110	10473	0.53	65	10417	0.00	47
19.2k	19.23k	0.16	64	19.20k	0.00	59	19.20k	0.00	35	19231	0.16	25
57.6k	56.82k	-1.36	21	57.60k	0.00	19	57.60k	0.00	11	55556	-3.55	8
115.2k	113.64k	-1.36	10	115.2k	0.00	9	115.2k	0.00	5	—	_	_

					SYNC	C = 0, BRGH	l = 1, BRC	G16 = 0					
BAUD	Fos	c = 4.00	0 MHz	Fosc	; = 3.686	4 MHz	Fos	c = 2.00	0 MHz	Fos	Fosc = 1.000 MHz		
RATE 300	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	
300	_	_		_	_	_	_	_	_	300	0.16	207	
1200	1202	0.16	207	1200	0.00	191	1202	0.16	103	1202	0.16	51	
2400	2404	0.16	103	2400	0.00	95	2404	0.16	51	2404	0.16	25	
9600	9615	0.16	25	9600	0.00	23	9615	0.16	12	—	_	_	
10417	10417	0.00	23	10473	0.53	21	10417	0.00	11	10417	0.00	5	
19.2k	19.23k	0.16	12	19.2k	0.00	11	—	_	_	—	_	_	
57.6k	_	_	_	57.60k	0.00	3	—	_	_	—	_	_	
115.2k	—	_		115.2k	0.00	1	_	_	_	—	_		

TABLE 12-5: BAUD RATES FOR ASYNCHRONOUS MODES (CONTINUED)

					SYNC	C = 0, BRG							
BAUD	Fosc	; = 20.00	0 MHz	Fosc = 18.432 MHz			Fosc	= 11.059	92 MHz	Fos	Fosc = 8.000 MHz		
RATE	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	
300	300.0	-0.01	4166	300.0	0.00	3839	300.0	0.00	2303	299.9	-0.02	1666	
1200	1200	-0.03	1041	1200	0.00	959	1200	0.00	575	1199	-0.08	416	
2400	2399	-0.03	520	2400	0.00	479	2400	0.00	287	2404	0.16	207	
9600	9615	0.16	129	9600	0.00	119	9600	0.00	71	9615	0.16	51	
10417	10417	0.00	119	10378	-0.37	110	10473	0.53	65	10417	0.00	47	
19.2k	19.23k	0.16	64	19.20k	0.00	59	19.20k	0.00	35	19.23k	0.16	25	
57.6k	56.818	-1.36	21	57.60k	0.00	19	57.60k	0.00	11	55556	-3.55	8	
115.2k	113.636	-1.36	10	115.2k	0.00	9	115.2k	0.00	5	_	_	_	

		SYNC = 0, BRGH = 0, BRG16 = 1													
BAUD	Fos	c = 4.00	0 MHz	Fosc	: = 3.686	4 MHz	Fos	c = 2.000) MHz	Fos	c = 1.000) MHz			
RATE 300	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)			
300	300.1	0.04	832	300.0	0.00	767	299.8	-0.108	416	300.5	0.16	207			
1200	1202	0.16	207	1200	0.00	191	1202	0.16	103	1202	0.16	51			
2400	2404	0.16	103	2400	0.00	95	2404	0.16	51	2404	0.16	25			
9600	9615	0.16	25	9600	0.00	23	9615	0.16	12	—	—	—			
10417	10417	0.00	23	10473	0.53	21	10417	0.00	11	10417	0.00	5			
19.2k	19.23k	0.16	12	19.20k	0.00	11	_	—	—	—	_	_			
57.6k	_	_	_	57.60k	0.00	3	—	_	_	—	_	_			
115.2k	—	_	_	115.2k	0.00	1	—	—	—	—	—	_			

				SYNC = 0	, BRGH	= 1, BRG16) = 1 or Sነ	/NC = 1,	BRG16 = 1			
BAUD	Fosc = 20.000 MHz			Fosc = 18.432 MHz			Fosc	= 11.059	92 MHz	Fosc = 8.000 MHz		
RATE	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)
300	300.0	0.00	16665	300.0	0.00	15359	300.0	0.00	9215	300.0	0.00	6666
1200	1200	-0.01	4166	1200	0.00	3839	1200	0.00	2303	1200	-0.02	1666
2400	2400	0.02	2082	2400	0.00	1919	2400	0.00	1151	2401	0.04	832
9600	9597	-0.03	520	9600	0.00	479	9600	0.00	287	9615	0.16	207
10417	10417	0.00	479	10425	0.08	441	10433	0.16	264	10417	0	191
19.2k	19.23k	0.16	259	19.20k	0.00	239	19.20k	0.00	143	19.23k	0.16	103
57.6k	57.47k	-0.22	86	57.60k	0.00	79	57.60k	0.00	47	57.14k	-0.79	34
115.2k	116.3k	0.94	42	115.2k	0.00	39	115.2k	0.00	23	117.6k	2.12	16

TABLE 12-5: BAUD RATES FOR ASYNCHRONOUS MODES (CONTINUED)

				SYNC = 0	, BRGH	= 1, BRG16	6 = 1 or SYNC = 1, BRG16 = 1						
BAUD	Fosc = 4.000 MHz			Fosc = 3.6864 MHz			Fos	c = 2.00) MHz	Fosc = 1.000 MHz			
RATE	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	
300	300.0	0.01	3332	300.0	0.00	3071	299.9	-0.02	1666	300.1	0.04	832	
1200	1200	0.04	832	1200	0.00	767	1199	-0.08	416	1202	0.16	207	
2400	2398	0.08	416	2400	0.00	383	2404	0.16	207	2404	0.16	103	
9600	9615	0.16	103	9600	0.00	95	9615	0.16	51	9615	0.16	25	
10417	10417	0.00	95	10473	0.53	87	10417	0.00	47	10417	0.00	23	
19.2k	19.23k	0.16	51	19.20k	0.00	47	19.23k	0.16	25	19.23k	0.16	12	
57.6k	58.82k	2.12	16	57.60k	0.00	15	55.56k	-3.55	8	—	—	_	
115.2k	111.1k	-3.55	8	115.2k	0.00	7	—	—	—	_	—	—	

12.3.1 AUTO-BAUD DETECT

The EUSART module supports automatic detection and calibration of the baud rate.

In the Auto-Baud Detect (ABD) mode, the clock to the BRG is reversed. Rather than the BRG clocking the incoming RX signal, the RX signal is timing the BRG. The Baud Rate Generator is used to time the period of a received 55h (ASCII "U") which is the Sync character for the LIN bus. The unique feature of this character is that it has five rising edges including the Stop bit edge.

Setting the ABDEN bit of the BAUDCTL register starts the auto-baud calibration sequence (Figure 12-6). While the ABD sequence takes place, the EUSART state machine is held in Idle. On the first rising edge of the receive line, after the Start bit, the SPBRG begins counting up using the BRG counter clock as shown in Table 12-6. The fifth rising edge will occur on the RX pin at the end of the eighth bit period. At that time, an accumulated value totaling the proper BRG period is left in SPBRGH, SPBRG register pair, the ABDEN bit is automatically cleared and the RCIF interrupt flag is set. The value in the RCREG needs to be read to clear the RCIF interrupt. RCREG content should be discarded. When calibrating for modes that do not use the SPBRGH register the user can verify that the SPBRG register did not overflow by checking for 00h in the SPBRGH register.

The BRG auto-baud clock is determined by the BRG16 and BRGH bits as shown in Table 12-6. During ABD, both the SPBRGH and SPBRG registers are used as a 16-bit counter, independent of the BRG16 bit setting. While calibrating the baud rate period, the SPBRGH and SPBRG registers are clocked at 1/8th the BRG base clock rate. The resulting byte measurement is the average bit time when clocked at full speed.

- Note 1: If the WUE bit is set with the ABDEN bit, auto-baud detection will occur on the byte following the Break character (see Section 12.3.2 "Auto-Wake-up on Break").
 - 2: It is up to the user to determine that the incoming character baud rate is within the range of the selected BRG clock source. Some combinations of oscillator frequency and EUSART baud rates are not possible due to bit error rates. Overall system timing and communication baud rates must be taken into consideration when using the Auto-Baud Detect feature.
 - **3:** After completion of the auto-baud sequence, the calculated auto-baud value will be the baud-rate plus 1.

TABLE 12-6: E	BRG COUNTER	CLOCK RATES
---------------	--------------------	--------------------

BRG16	BRGH	BRG Base Clock	BRG ABD Clock
0	0	Fosc/64	Fosc/512
0	1	Fosc/16	Fosc/128
1	0	Fosc/16	Fosc/128
1	1	Fosc/4	Fosc/32

Note: During the ABD sequence, SPBRG and SPBRGH registers are both used as a 16-bit counter, independent of BRG16 setting.

FIGURE 12-6: AUTOMATIC BAUD RATE CALIBRATION

	Edge #1	← Edge #5 Stop bit
		I
-		
r — ,	l – – – – – – – – – – – – – – – – – – –	Auto Cleared
•		<u> </u>
i		
	<u>.</u>	
		<u> </u>
l l	1	(
l l		h
l I		
	XXh	1Ch
•	XXh	00h
		XXh XXh

12.3.2 AUTO-WAKE-UP ON BREAK

During Sleep mode, all clocks to the EUSART are suspended. Because of this, the Baud Rate Generator is inactive and a proper character reception cannot be performed. The Auto-Wake-up feature allows the controller to wake-up due to activity on the RX/DT line. This feature is available only in Asynchronous mode.

The Auto-Wake-up feature is enabled by setting the WUE bit of the BAUDCTL register. Once set, the normal receive sequence on RX/DT is disabled and the EUSART remains in an Idle state, monitoring for a wake-up event independent of the CPU mode. A wake-up event consists of a high-to-low transition on the RX/DT line. (This coincides with the start of a Sync Break or a wake-up signal character for the LIN protocol.)

The EUSART module generates an RCIF interrupt coincident with the wake-up event. The interrupt is generated synchronously to the Q clocks in normal CPU operating modes (Figure 12-7), and asynchronously if the device is in Sleep mode (Figure 12-8). The interrupt condition is cleared by reading the RCREG register.

The WUE bit is automatically cleared by the low-to-high transition on the RX line at the end of the Break. This signals to the user that the Break event is over. At this point, the EUSART module is in Idle mode waiting to receive the next character.

12.3.2.1 Special Considerations

Break Character

To avoid character errors or character fragments during a wake-up event, the wake-up character must be all zeros.

When the wake-up is enabled the function works independent of the low time on the data stream. If the WUE bit is set and a valid non-zero character is received, the low time from the Start bit to the first rising edge will be interpreted as the wake-up event. The remaining bits in the character will be received as a fragmented character and subsequent characters can result in framing or overrun errors.

Therefore, the initial character in the transmission must be all '0's. This must be 10 or more bit times, 13-bit times recommended for LIN bus, or any number of bit times for standard RS-232 devices.

Oscillator Startup Time

Oscillator start-up time must be considered, especially in applications using oscillators with longer start-up intervals (i.e., LP, XT or HS/PLL mode). The Sync Break (or wake-up signal) character must be of sufficient length, and be followed by a sufficient interval, to allow enough time for the selected oscillator to start and provide proper initialization of the EUSART.

WUE Bit

The wake-up event causes a receive interrupt by setting the RCIF bit. The WUE bit is cleared in hardware by a rising edge on RX/DT. The interrupt condition is then cleared in software by reading the RCREG register and discarding its contents.

To ensure that no actual data is lost, check the RCIDL bit to verify that a receive operation is not in process before setting the WUE bit. If a receive operation is not occurring, the WUE bit may then be set just prior to entering the Sleep mode.

FIGURE 12-7: AUTO-WAKE-UP BIT (WUE) TIMING DURING NORMAL OPERATION







2: The EUSART remains in Idle while the WUE bit is set.

12.3.3 BREAK CHARACTER SEQUENCE

The EUSART module has the capability of sending the special Break character sequences that are required by the LIN bus standard. A Break character consists of a Start bit, followed by 12 '0' bits and a Stop bit.

To send a Break character, set the SENDB and TXEN bits of the TXSTA register. The Break character transmission is then initiated by a write to the TXREG. The value of data written to TXREG will be ignored and all '0's will be transmitted.

The SENDB bit is automatically reset by hardware after the corresponding Stop bit is sent. This allows the user to preload the transmit FIFO with the next transmit byte following the Break character (typically, the Sync character in the LIN specification).

The TRMT bit of the TXSTA register indicates when the transmit operation is active or idle, just as it does during normal transmission. See Figure 12-9 for the timing of the Break character sequence.

12.3.3.1 Break and Sync Transmit Sequence

The following sequence will start a message frame header made up of a Break, followed by an auto-baud Sync byte. This sequence is typical of a LIN bus master.

- 1. Configure the EUSART for the desired mode.
- 2. Set the TXEN and SENDB bits to enable the Break sequence.
- 3. Load the TXREG with a dummy character to initiate transmission (the value is ignored).
- 4. Write '55h' to TXREG to load the Sync character into the transmit FIFO buffer.
- 5. After the Break has been sent, the SENDB bit is reset by hardware and the Sync character is then transmitted.

When the TXREG becomes empty, as indicated by the TXIF, the next data byte can be written to TXREG.

12.3.4 RECEIVING A BREAK CHARACTER

The Enhanced EUSART module can receive a Break character in two ways.

The first method to detect a Break character uses the FERR bit of the RCSTA register and the Received data as indicated by RCREG. The Baud Rate Generator is assumed to have been initialized to the expected baud rate.

A Break character has been received when:

- RCIF bit is set
- FERR bit is set
- RCREG = 00h

The second method uses the Auto-Wake-up feature described in **Section 12.3.2** "**Auto-Wake-up on Break**". By enabling this feature, the EUSART will sample the next two transitions on RX/DT, cause an RCIF interrupt, and receive the next data byte followed by another interrupt.

Note that following a Break character, the user will typically want to enable the Auto-Baud Detect feature. For both methods, the user can set the ABDEN bit of the BAUDCTL register before placing the EUSART in Sleep mode.



FIGURE 12-9: SEND BREAK CHARACTER SEQUENCE

12.4 EUSART Synchronous Mode

Synchronous serial communications are typically used in systems with a single master and one or more slaves. The master device contains the necessary circuitry for baud rate generation and supplies the clock for all devices in the system. Slave devices can take advantage of the master clock by eliminating the internal clock generation circuitry.

There are two signal lines in Synchronous mode: a bidirectional data line and a clock line. Slaves use the external clock supplied by the master to shift the serial data into and out of their respective receive and transmit shift registers. Since the data line is bidirectional, synchronous operation is half-duplex only. Half-duplex refers to the fact that master and slave devices can receive and transmit data, but not both simultaneously. The EUSART can operate as either a master or slave device.

Start and Stop bits are not used in synchronous transmissions.

12.4.1 SYNCHRONOUS MASTER MODE

The following bits are used to configure the EUSART for Synchronous Master operation:

- SYNC = 1
- CSRC = 1
- SREN = 0 (for transmit); SREN = 1 (for receive)
- CREN = 0 (for transmit); CREN = 1 (for receive)
- SPEN = 1

Setting the SYNC bit of the TXSTA register configures the device for synchronous operation. Setting the CSRC bit of the TXSTA register configures the device as a master. Clearing the SREN and CREN bits of the RCSTA register ensures that the device is in the Transmit mode, otherwise the device will be configured to receive. Setting the SPEN bit of the RCSTA register enables the EUSART. If the RX/DT or TX/CK pins are shared with an analog peripheral the analog I/O functions must be disabled by clearing the corresponding ANSEL bits.

12.4.1.1 Master Clock

Synchronous data transfers use a separate clock line, which is synchronous with the data. A device configured as a master transmits the clock on the TX/CK line. The TX/CK pin is automatically configured as an output when the EUSART is configured for synchronous transmit operation. Serial data bits change on the leading edge to ensure they are valid at the trailing edge of each clock. One clock cycle is generated for each data bit. Only as many clock cycles are generated as there are data bits.

12.4.1.2 Clock Polarity

A clock polarity option is provided for Microwire compatability. Clock polarity is selected with the SCKP bit of the BAUDCTL register. Setting the SCKP bit sets

the clock Idle state as high. When the SCKP bit is set, the data changes on the falling edge of each clock. Clearing the SCKP bit sets the Idle state as low. When the SCKP bit is cleared, the data changes on the rising edge of each clock.

12.4.1.3 Synchronous Master Transmission

Data is transferred out of the device on the RX/DT pin. The RX/DT and TX/CK pin output drivers are automatically enabled when the EUSART is configured for synchronous master transmit operation.

A transmission is initiated by writing a character to the TXREG register. If the TSR still contains all or part of a previous character the new character data is held in the TXREG until the last bit of the previous character has been transmitted. If this is the first character, or the previous character has been completely flushed from the TSR, the data in the TXREG is immediately transferred to the TSR. The transmission of the character commences immediately following the transfer of the data to the TSR from the TXREG.

Each data bit changes on the leading edge of the master clock and remains valid until the subsequent leading clock edge.

Note: The TSR register is not mapped in data memory, so it is not available to the user.

- 12.4.1.4 Synchronous Master Transmission Set-up:
- 1. Initialize the SPBRGH, SPBRG register pair and the BRGH and BRG16 bits to achieve the desired baud rate (see Section 12.3 "EUSART Baud Rate Generator (BRG)").
- 2. Enable the synchronous master serial port by setting bits SYNC, SPEN and CSRC.
- 3. Disable Receive mode by clearing bits SREN and CREN.
- 4. Enable Transmit mode by setting the TXEN bit.
- 5. If 9-bit transmission is desired, set the TX9 bit.
- 6. If interrupts are desired, set the TXIE, GIE and PEIE interrupt enable bits.
- 7. If 9-bit transmission is selected, the ninth bit should be loaded in the TX9D bit.
- 8. Start transmission by loading data to the TXREG register.



FIGURE 12-10: SYNCHRONOUS TRANSMISSION

FIGURE 12-11: SYNCHRONOUS TRANSMISSION (THROUGH TXEN)



TABLE 12-7: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
BAUDCTL	ABDOVF	RCIDL	_	SCKP	BRG16	—	WUE	ABDEN	01-0 0-00	01-0 0-00
INTCON	GIE	PEIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF	x000 0000	x000 0000
PIE1	—	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	-000 0000	-000 0000
PIR1	—	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	-000 0000	-000 0000
RCREG	EUSART	Receive Da	ita Register						0000 0000	0000 0000
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	x000 0000	x000 0000
SPBRG	BRG7	BRG6	BRG5	BRG4	BRG3	BRG2	BRG1	BRG0	0000 0000	0000 0000
SPBRGH	BRG15	BRG14	BRG13	BRG12	BRG11	BRG10	BRG9	BRG8	0000 0000	0000 0000
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111
TXREG	EUSART	Transmit Da	ata Registe	r					0000 0000	0000 0000
TXSTA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	0000 0010	0000 0010

Legend: x = unknown, - = unimplemented read as '0'. Shaded cells are not used for Synchronous Master Transmission.

12.4.1.5 Synchronous Master Reception

Data is received at the RX/DT pin. The RX/DT and TX/ CK pin output drivers are automatically disabled when the EUSART is configured for synchronous master receive operation.

In Synchronous mode, reception is enabled by setting either the Single Receive Enable bit (SREN of the RCSTA register) or the Continuous Receive Enable bit (CREN of the RCSTA register).

When SREN is set and CREN is clear, only as many clock cycles are generated as there are data bits in a single character. The SREN bit is automatically cleared at the completion of one character. When CREN is set, clocks are continuously generated until CREN is cleared. If CREN is cleared in the middle of a character the CK clock stops immediately and the partial character is discarded. If SREN and CREN are both set, then SREN is cleared at the completion of the first character and CREN takes precedence.

To initiate reception, set either SREN or CREN. Data is sampled at the RX/DT pin on the trailing edge of the TX/CK clock pin and is shifted into the Receive Shift Register (RSR). When a complete character is received into the RSR, the RCIF bit is set and the character is automatically transferred to the two character receive FIFO. The Least Significant eight bits of the top character in the receive FIFO are available in RCREG. The RCIF bit remains set as long as there are un-read characters in the receive FIFO.

12.4.1.6 Receive Overrun Error

The receive FIFO buffer can hold two characters. An overrun error will be generated if a third character, in its entirety, is received before RCREG is read to access the FIFO. When this happens the OERR bit of the RCSTA register is set. Previous data in the FIFO will not be overwritten. The two characters in the FIFO buffer can be read, however, no additional characters will be received until the error is cleared. The OERR bit can only be cleared by clearing the overrun condition. If the overrun error occurred when the SREN bit is set and CREN is clear then the error is cleared by reading RCREG. If the overrun occurred when the CREN bit is set then the error condition is cleared by either clearing the CREN bit of the RCSTA register or by clearing the SPEN bit which resets the EUSART.

12.4.1.7 Receiving 9-bit Characters

The EUSART supports 9-bit character reception. When the RX9 bit of the RCSTA register is set the EUSART will shift 9-bits into the RSR for each character received. The RX9D bit of the RCSTA register is the ninth, and Most Significant, data bit of the top unread character in the receive FIFO. When reading 9-bit data from the receive FIFO buffer, the RX9D data bit must be read before reading the 8 Least Significant bits from the RCREG.

- 12.4.1.8 Synchronous Master Reception Setup:
- 1. Initialize the SPBRGH, SPBRG register pair for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
- 2. Enable the synchronous master serial port by setting bits SYNC, SPEN and CSRC.
- 3. Ensure bits CREN and SREN are clear.
- 4. If using interrupts, set the GIE and PEIE bits of the INTCON register and set RCIE.
- 5. If 9-bit reception is desired, set bit RX9.
- 6. Start reception by setting the SREN bit or for continuous reception, set the CREN bit.
- 7. Interrupt flag bit RCIF will be set when reception of a character is complete. An interrupt will be generated if the enable bit RCIE was set.
- 8. Read the RCSTA register to get the ninth bit (if enabled) and determine if any error occurred during reception.
- 9. Read the 8-bit received data by reading the RCREG register.
- 10. If an overrun error occurs, clear the error by either clearing the CREN bit of the RCSTA register or by clearing the SPEN bit which resets the EUSART.

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RX/DT	
TX/CK pin	
SREN bit	'0'
RCIF bit (Interrupt) Read	
RXREG Note: Timing diagram demonstrates Sync Master mode with bit SREN = 1 and bit BRGH = 0.	

FIGURE 12-12: SYNCHRONOUS RECEPTION (MASTER MODE, SREN)

TABLE 12-8: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
BAUDCTL	ABDOVF	RCIDL	—	SCKP	BRG16	—	WUE	ABDEN	01-0 0-00	01-0 0-00
INTCON	GIE	PEIE	TOIE	INTE	RBIE	T0IF	INTF	RBIF	x000 000x	x000 000x
PIE1	—	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	-000 0000	-000 0000
PIR1	—	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	-000 0000	-000 0000
RCREG	EUSART	Receive Da	ta Register						0000 0000	0000 0000
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	x000 000x	x000 000x
SPBRG	BRG7	BRG6	BRG5	BRG4	BRG3	BRG2	BRG1	BRG0	0000 0000	0000 0000
SPBRGH	BRG15	BRG14	BRG13	BRG12	BRG11	BRG10	BRG9	BRG8	0000 0000	0000 0000
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111
TXREG	EUSART Transmit Data Register								0000 0000	0000 0000
TXSTA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	0000 0010	0000 0010

Legend: x = unknown, - = unimplemented read as '0'. Shaded cells are not used for Synchronous Master Reception.

12.4.2 SYNCHRONOUS SLAVE MODE

The following bits are used to configure the EUSART for Synchronous slave operation:

- SYNC = 1
- CSRC = 0
- SREN = 0 (for transmit); SREN = 1 (for receive)
- CREN = 0 (for transmit); CREN = 1 (for receive)
- SPEN = 1

Setting the SYNC bit of the TXSTA register configures the device for synchronous operation. Clearing the CSRC bit of the TXSTA register configures the device as a slave. Clearing the SREN and CREN bits of the RCSTA register ensures that the device is in the Transmit mode, otherwise the device will be configured to receive. Setting the SPEN bit of the RCSTA register enables the EUSART. If the RX/DT or TX/CK pins are shared with an analog peripheral the analog I/O functions must be disabled by clearing the corresponding ANSEL bits.

12.4.2.1 EUSART Synchronous Slave Transmit

The operation of the Synchronous Master and Slave modes are identical (see **Section 12.4.1.3 "Synchronous Master Transmission")**, except in the case of the Sleep mode. If two words are written to the TXREG and then the SLEEP instruction is executed, the following will occur:

- 1. The first character will immediately transfer to the TSR register and transmit.
- 2. The second word will remain in TXREG register.
- 3. The TXIF bit will not be set.
- After the first character has been shifted out of TSR, the TXREG register will transfer the second character to the TSR and the TXIF bit will now be set.
- 5. If the PEIE and TXIE bits are set, the interrupt will wake the device from Sleep and execute the next instruction. If the GIE bit is also set, the program will call the interrupt service routine.
- 12.4.2.2 Synchronous Slave Transmission Set-up:
- 1. Set the SYNC and SPEN bits and clear the CSRC bit.
- 2. Clear the CREN and SREN bits.
- 3. If using interrupts, ensure that the GIE and PEIE bits of the INTCON register are set and set the TXIE bit.
- 4. If 9-bit transmission is desired, set the TX9 bit.
- 5. Enable transmission by setting the TXEN bit.
- 6. If 9-bit transmission is selected, insert the Most Significant bit into the TX9D bit.
- 7. Start transmission by writing the Least Significant 8 bits to the TXREG register.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
BAUDCTL	ABDOVF	RCIDL	_	SCKP	BRG16	_	WUE	ABDEN	01-0 0-00	01-0 0-00
INTCON	GIE	PEIE	TOIE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	0000 000x
PIE1	_	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	-000 0000	-000 0000
PIR1	_	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	-000 0000	-000 0000
RCREG	EUSART R	Receive Da	ta Register						0000 0000	0000 0000
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	x000 000x	0000 000x
SPBRG	BRG7	BRG6	BRG5	BRG4	BRG3	BRG2	BRG1	BRG0	0000 0000	0000 0000
SPBRGH	BRG15	BRG14	BRG13	BRG12	BRG11	BRG10	BRG9	BRG8	0000 0000	0000 0000
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111
TXREG	EUSART Transmit Data Register								0000 0000	0000 0000
TXSTA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	0000 0010	0000 0010

TABLE 12-9: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE TRANSMISSION

Legend: x = unknown, - = unimplemented read as '0'. Shaded cells are not used for Synchronous Slave Transmission.

12.4.2.3 **EUSART Synchronous Slave** Reception

The operation of the Synchronous Master and Slave modes is identical (Section 12.4.1.5 "Synchronous Master Reception"), with the following exceptions:

- Sleep
- · CREN bit is always set, therefore the receiver is never idle
- · SREN bit, which is a "don't care" in Slave mode

A character may be received while in Sleep mode by setting the CREN bit prior to entering Sleep. Once the word is received, the RSR register will transfer the data to the RCREG register. If the RCIE enable bit is set, the interrupt generated will wake the device from Sleep and execute the next instruction. If the GIE bit is also set, the program will branch to the interrupt vector.

- 12.4.2.4 Synchronous Slave Reception Setup:
- 1. Set the SYNC and SPEN bits and clear the CSRC bit.
- 2. If using interrupts, ensure that the GIE and PEIE bits of the INTCON register are set and set the RCIE bit.
- If 9-bit reception is desired, set the RX9 bit. 3.
- Set the CREN bit to enable reception. 4.
- 5. The RCIF bit will be set when reception is complete. An interrupt will be generated if the RCIE bit was set.
- If 9-bit mode is enabled, retrieve the Most 6. Significant bit from the RX9D bit of the RCSTA register.
- Retrieve the 8 Least Significant bits from the 7. receive FIFO by reading the RCREG register.
- If an overrun error occurs, clear the error by 8. either clearing the CREN bit of the RCSTA register or by clearing the SPEN bit which resets the EUSART.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
BAUDCTL	ABDOVF	RCIDL	—	SCKP	BRG16	_	WUE	ABDEN	01-0 0-00	01-0 0-00
INTCON	GIE	PEIE	TOIE	INTE	RBIE	T0IF	INTF	RBIF	x000 000x	0000 000x
PIE1	—	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	-000 0000	-000 0000
PIR1	—	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	-000 0000	-000 0000
RCREG	EUSART I	Receive Da	ta Register						0000 0000	0000 0000
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	x000 000x	0000 000x
SPBRG	BRG7	BRG6	BRG5	BRG4	BRG3	BRG2	BRG1	BRG0	0000 0000	0000 0000
SPBRGH	BRG15	BRG14	BRG13	BRG12	BRG11	BRG10	BRG9	BRG8	0000 0000	0000 0000
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111
TXREG	EUSART Transmit Data Register								0000 0000	0000 0000
TXSTA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	0000 0010	0000 0010

TABLE 12-10: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE RECEPTION

Legend: x = unknown, - = unimplemented read as '0'. Shaded cells are not used for Synchronous Slave Reception.

13.0 MASTER SYNCHRONOUS SERIAL PORT (MSSP) MODULE

13.1 Master SSP (MSSP) Module Overview

The Master Synchronous Serial Port (MSSP) module is a serial interface useful for communicating with other peripheral or microcontroller devices. These peripheral devices may be Serial EEPROMs, shift registers, display drivers, A/D converters, etc. The MSSP module can operate in one of two modes:

- Serial Peripheral Interface (SPI)
- Inter-Integrated Circuit[™] (I²C[™])
 - Full Master mode
 - Slave mode (with general address call).

The ${\rm I}^2{\rm C}$ interface supports the following modes in hardware:

- Master mode
- Multi-Master mode
- Slave mode.

13.2 Control Registers

The MSSP module has three associated registers. These include a STATUS register and two control registers.

Register 13-1 shows the MSSP STATUS register (SSPSTAT), Register 13-2 shows the MSSP Control Register 1 (SSPCON), and Register 13-3 shows the MSSP Control Register 2 (SSPCON2).

REGISTER 13-1: SSPSTAT: SSP STATUS REGISTER

R/W-0	R/W-0	R-0	R-0	R-0	R-0	R-0	R-0
SMP	CKE	D/Ā	Р	S	R/W	UA	BF
bit 7	•						bit 0
Legend:							
R = Readable bit		W = Writable bit		U = Unimplem	ented bit, read as	0'	
-n = Value at PO	R	'1' = Bit is set		'0' = Bit is clea	red	x = Bit is unknov	vn
bit 7	SMP: Sample b	bit					
	1 = Input data s	ampled at end of a	data output tim	e			
	0 = Input data s	ampled at middle	of data output	time			
	SMP must be c	<u>e:</u> leared when SPI is	s used in Slave	mode			
	In I ² C Master of	r Slave mode:					
	1 = Slew rate o	control disabled for	standard speed	ed mode (100 kH ode (400 kHz)	lz and 1 MHz)		
bit 6	CKE: SPI Clock	k Edge Select bit	ingri opood int				
	<u>CKP = 0:</u>						
	1 = Data transm	nitted on rising edg	le of SCK				
	CKP = 1:						
	1 = Data transm	nitted on falling edg	ge of SCK				
bit 5	0 = Data transm	$\frac{1}{10000}$ bit (1 ² C mode)	e of SCK				
DIL 5	1 = Indicates th	at the last byte rec	eived or transi	mitted was data			
	0 = Indicates th	at the last byte rec	eived or transi	mitted was addre	ess		
bit 4	P: Stop bit						
	(I ⁺ C mode only. 1 = Indicates th	at a Stop bit has b	when the MS: een detected la	SP module is dis ast (this bit is '0'	abled, SSPEN is c on Reset)	leared.)	
	0 = Stop bit was	s not detected last					
bit 3	S: Start bit						
	(I ² C mode only.	This bit is cleared	when the MSS	SP module is dis	abled, SSPEN is c	leared.)	
	0 = Start bit was	s not detected last			on Reset)		
bit 2	R/W: Read/Writ	te bit information (I	² C mode only)	1			
	This bit holds th the next Start bi	e R/W bit informa <u>t</u> it. Stop bit. or not A	<u>ion f</u> ollowing th ACK bit.	ne last address m	natch. This bit is on	ly valid from the a	ddress match to
	In I ² C Slave mo	ode:					
	1 = Read 0 = Write						
	In I ² C Master m	node:					
	1 = Transmit is	s in progress					
	OR-ing th	is bit with SEN, RS	SEN, PEN, RC	EN, or ACKEN w	vill indicate if the M	SSP is in Idle mod	le.
bit 1	UA: Update Ad	dress bit (10-bit I ²	C mode only)				
	1 = Indicates th 0 = Address do	at the user needs es not need to be	to update the a updated	address in the SS	SPADD register		
bit 0	BF: Buffer Full	Status bit					
	<u>Receive (SPI a</u>	nd I ² C modes):					
	1 = Receive cor	mplete, SSPBUF is	s full IF is emoty				
	Transmit (I ² C m	node only):					
	1 = Data transm	nit in progress (doe	es not include t	he ACK and Sto	p bits), SSPBUF is	full	
	0 = Data transm	nit complete (does	not include the	e ACK and Stop	DITS), SSPBUF is e	mpty	

REGISTER 13-2: SSPCON: SSP CONTROL REGISTER 1

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0
bit 7	•	· · · · · · · · · · · · · · · · · · ·		•	·		bit 0
Legend:							
R = Readable bit		W = Writable bit		U = Unimpleme	ented bit, read as '0'		
-n = Value at POR	1	'1' = Bit is set		'0' = Bit is clear	ed	x = Bit is unknow	wn
bit 7	WCOL: Write C Master mode: 1 = A write to t	ollision Detect bit he SSPBUF registe	er was attempte	ed while the I ² C co	nditions were not va	alid for a transmis	sion to be started
	<u>Slave mode:</u> 1 = The SSPBI 0 = No collisio	UF register is written n	while it is still tr	ansmitting the previ	ous word (must be c	leared in software))
bit 6	SSPOV: Receiv In SPI mode: 1 = A new byte is lost. Ove data, to avv initiated by 0 = No overflo	e Overflow Indicato is received while the erflow can only occur oid setting overflow. writing to the SSPB w	or bit e SSPBUF regi r in Slave mode In Master mode UF register (mo	ster is still holding th In Slave mode, th I, the overflow bit is ust be cleared in so	ne previous data. In o le user must read the s not set since each r ftware).	case of overflow, to SSPBUF, even i new reception (an	ne data in SSPSR f only transmitting d transmission) is
	$\frac{\ln l^2 C \text{ mode:}}{1 = A \text{ byte is remove (must)}}$ $0 = \text{ No overflow}$	eceived while the S st be cleared in sof w	SPBUF regist tware).	er is still holding th	ne previous byte. S	SPOV is a "don't	care" in Transmit
bit 5	SSPEN: Synchr In both modes, $1 = \text{Enables se}$ 0 = Disables s 1 = Enables se 1 = Enables the 0 = Disables s	ronous Serial Port E when enabled, thes erial port and configu- erial port and config- e serial port and config-	Enable bit se pins must be ures SCK, SDO gures these pir figures the SDA	e properly configur , SDI and \overline{SS} as the ns as I/O port pins A and SCL pins as t	red as input or output e source of the seria the source of the seri	ut I port pins al port pins	
bit 4	CKP: Clock Pol. In SPI mode: 1 = Idle state for 0 = Idle state for In I ² C Slave mo SCK release co 1 = Enable clock 0 = Holds clock In I ² C Master m Unused in this n	arity Select bit r clock is a high lev r clock is a low leve de: ntrol k low (clock stretch). ode: node	el I (Used to ensu	ire data setup time	s.)		
bit 3-0	$\begin{array}{l} \text{SSPM<3:0>: Sy}\\ 0000 = \text{SPI Mas}\\ 0001 = \text{SPI Mas}\\ 0010 = \text{SPI Mas}\\ 0011 = \text{SPI Mas}\\ 0100 = \text{SPI Slav}\\ 0101 = \text{SPI Slav}\\ 0101 = \text{SPI Slav}\\ 0101 = \text{I}^2\text{C Slav}\\ 0101 = \text{I}^2\text{C Slav}\\ 1000 = \text{I}^2\text{C Mas}\\ 1010 = \text{Reserve}\\ 1011 = \text{I}^2\text{C Slav}\\ 1100 = \text{Reserve}\\ 1101 = \text{Reserve}\\ 1110 = \text{I}^2\text{C Slav}\\ 1111 $	vnchronous Serial F ster mode, clock = I ster mode, clock = I ster mode, clock = S ve mode, clock = F ask function ed ware controlled Ma ed ve mode, 7-bit addr ve mode, 7-bit addr ve mode, 7-bit addr	Port Mode Sele Fosc/4 Fosc/64 TMR2 o <u>utp</u> ut/2 CK pin, <u>SS</u> pin CK pin, <u>SS</u> pin ess Irress Fosc / (4 * (SS ster mode (Sla ess with Start a Irress with Start	control enabled control disabled, PADD+1)) we idle) and Stop bit interru	SS can be used as upts enabled rupts enabled	I/O pin	

R/W-0) R-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	
GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	
bit 7 bit 0								
Legend:								
R = Readable bit		W = Writable bit		U = Unimplemented bit, read as '0'				
-n = Value at POR		'1' = Bit is set		'0' = Bit is cleared x = Bit is unknown				
bit 7	GCEN: Generation Generation 1 = Enable int	GCEN: General Call Enable bit (in I ² C Slave mode only) 1 = Enable interrupt when a general call address (0000h) is received in the SSPSR						
	0 = General call address disabled							
bit 6	ACKSTAT: Acknowledge Status bit (in I ² C Master mode only)							
	In Master Transmit mode:							
	 a Acknowledge was not received from slave a Acknowledge was received from slave 							
bit 5	ACKDT: Acknowledge Data bit (in l^2 C Master mode only)							
	In Master Receive mode:							
	Value transmitted when the user initiates an Acknowledge sequence at the end of a receive							
	1 = Not Acknowledge							
bit 1	0 = Acknowledge							
DIL 4	In Master Receive mode:							
	1 = Initiate Acknowledge sequence on SDA and SCL pins, and transmit ACKDT data bit							
	Automatically cleared by hardware. 0 = Acknowledge sequence idle							
bit 3	RCEN: Receive Enable bit (in I ² C Master mode only)							
	$1 = \text{Enables Receive mode for } I^2 C$							
	0 = Receive id	dle						
bit 2 PEN: Stop Condition Enable bit (in I ² C Master mode only)								
	<u>SCK Release</u>	<u>Control:</u>	SDA and SC	l nine Autom	atically cleared	by bardware		
	0 = Stop cond	lition Idle			alleally cleared	by hardware.		
bit 1	RSEN: Repea	RSEN: Repeated Start Condition Enabled bit (in I ² C Master mode only)						
	1 = Initiate Ro 0 = Repeated	epeated Start o d Start conditio	condition on S n Idle	DA and SCL p	ins. Automatica	lly cleared by h	nardware.	
bit 0 SEN: Start Condition Enabled bit (in I ² C Master mode only)								
	1 = Initiate Sta 0 = Start conc	art condition or dition Idle	n SDA and SC	L pins. Autom	atically cleared	by hardware.		
Note 1:	For bits ACKEN, R	CEN, PEN, R	SEN, SEN: If t	he I ² C module	is not in the Idl	e mode, this bi	t may not be	

REGISTER 13-3: SSPCON2: SSP CONTROL REGISTER 2

Note 1: For bits ACKEN, RCEN, PEN, RSEN, SEN: If the I²C module is not in the Idle mode, this bit may not be set (no spooling) and the SSPBUF may not be written (or writes to the SSPBUF are disabled).
13.3 SPI Mode

The SPI mode allows 8 bits of data to be synchronously transmitted and received, simultaneously. All four modes of SPI are supported. To accomplish communication, typically three pins are used:

- Serial Data Out (SDO) RC5/SDO
- Serial Data In (SDI) RC4/SDI/SDA
- Serial Clock (SCK) RC3/SCK/SCL

Additionally, a fourth pin may be used when in any Slave mode of operation:

• Slave Select (SS) - RA5/SS/AN4

13.3.1 OPERATION

When initializing the SPI, several options need to be specified. This is done by programming the appropriate control bits SSPCON<5:0> and SSPSTAT<7:6>. These control bits allow the following to be specified:

- Master mode (SCK is the clock output)
- Slave mode (SCK is the clock input)
- Clock polarity (Idle state of SCK)
- Data input sample phase (middle or end of data output time)
- Clock edge (output data on rising/falling edge of SCK)
- Clock rate (Master mode only)
- Slave Select mode (Slave mode only)

Figure 13-1 shows the block diagram of the MSSP module, when in SPI mode.

FIGURE 13-1:

MSSP BLOCK DIAGRAM (SPI MODE)



Note: I/O pins have diode protection to VDD and VSS.

The MSSP consists of a transmit/receive shift register (SSPSR) and a buffer register (SSPBUF). The SSPSR shifts the data in and out of the device, MSb first. The SSPBUF holds the data that was written to the SSPSR, until the received data is ready. Once the 8 bits of data have been received, that byte is moved to the SSPBUF register. Then, the buffer full-detect bit BF of the SSP-STAT register and the interrupt flag bit SSPIF of the PIR1 register are set. This double buffering of the received data (SSPBUF) allows the next byte to start reception before reading the data that was just received. Any write to the SSPBUF register during transmission/reception of data will be ignored, and the write collision detect bit WCOL of the SSPCON register will be set. User software must clear the WCOL bit so that it can be determined if the following write(s) to the SSPBUF register completed successfully.

When the application software is expecting to receive valid data, the SSPBUF should be read before the next byte of data to transfer is written to the SSPBUF. The buffer full bit BF of the SSPSTAT register indicates when SSPBUF has been loaded with the received data (transmission is complete). When the SSPBUF is read, the BF bit is cleared. This data may be irrelevant if the SPI is only a transmitter. Generally, the MSSP Interrupt is used to determine when the transmission/reception has completed. The SSPBUF must be read and/or written. If the interrupt method is not going to be used, then software polling can be done to ensure that a write collision does not occur. Example 13-1 shows the loading of the SSPBUF (SSPSR) for data transmission.

The SSPSR is not directly readable or writable, and can only be accessed by addressing the SSPBUF register. Additionally, the MSSP STATUS register (SSPSTAT register) indicates the various status conditions.

13.3.2 ENABLING SPI I/O

To enable the serial port, SSP Enable bit SSPEN of the SSPCON register must be set. To reset or reconfigure SPI mode, clear the SSPEN bit, re-initialize the SSPCON registers, and then set the SSPEN bit. This configures the SDI, SDO, SCK and SS pins as serial port pins. For the pins to behave as the serial port function, some must have their data direction bits (in the TRIS register) appropriately programmed. That is:

- SDI is automatically controlled by the SPI module
- SDO must have TRISC<5> bit cleared
- SCK (Master mode) must have TRISC<3> bit cleared
- SCK (Slave mode) must have TRISC<3> bit set
- SS must have TRISA<5> bit set

Any serial port function that is not desired may be overridden by programming the corresponding data direction (TRIS) register to the opposite value.

EXAMPLE 13-1: LOADING THE SSPBUF (SSPSR) REGISTER

s data been received (transmit complete)?
EG reg = contents of SSPBUF
ve in user RAM, if data is meaningful
reg = contents of TXDATA
w data to xmit
s r w

13.3.3 MASTER MODE

The master can initiate the data transfer at any time because it controls the SCK. The master determines when the slave is to broadcast data by the software protocol.

In Master mode, the data is transmitted/received as soon as the SSPBUF register is written to. If the SPI is only going to receive, the SDO output could be disabled (programmed as an input). The SSPSR register will continue to shift in the signal present on the SDI pin at the programmed clock rate. As each byte is received, it will be loaded into the SSPBUF register as a normal received byte (interrupts and Status bits appropriately set). This could be useful in receiver applications as a "Line Activity Monitor" mode. The clock polarity is selected by appropriately programming the CKP bit of the SSPCON register. This, then, would give waveforms for SPI communication as shown in Figure 13-2, Figure 13-4 and Figure 13-5, where the MSb is transmitted first. In Master mode, the SPI clock rate (bit rate) is user programmable to be one of the following:

- Fosc/4 (or Tcy)
- Fosc/16 (or 4 Tcy)
- Fosc/64 (or 16 Tcy)
- Timer2 output/2

This allows a maximum data rate (at 40 MHz) of 10.00 Mbps.

Figure 13-2 shows the waveforms for Master mode. When the CKE bit of the SSPSTAT register is set, the SDO data is valid before there is a clock edge on SCK. The change of the input sample is shown based on the state of the SMP bit of the SSPSTAT register. The time when the SSPBUF is loaded with the received data is shown.





13.3.4 SLAVE MODE

In Slave mode, the data is transmitted and received as the external clock pulses appear on SCK. When the last bit is latched, the SSPIF interrupt flag bit of the PIR1 register is set.

While in Slave mode, the external clock is supplied by the external clock source on the SCK pin. This external clock must meet the minimum high and low times, as specified in the electrical specifications.

While in Sleep mode, the slave can transmit/receive data. When a byte is received, the device will wake-up from Sleep.

13.3.5 SLAVE SELECT SYNCHRONIZATION

The \overline{SS} pin allows a Synchronous Slave mode. The SPI must be in Slave mode with \overline{SS} pin control enabled (SSPCON<3:0> = 04h). The pin must not be driven low for the \overline{SS} pin to function as an input. The Data Latch must be high. When the \overline{SS} pin is low, transmission and reception are enabled and the SDO pin is driven. When the \overline{SS} pin goes high,

the SDO pin is no longer driven, even if in the middle of a transmitted byte, and becomes a floating output. External pull-up/pull-down resistors may be desirable, depending on the application.

Note	1: When the SPI is in Slave mode with \overline{SS}
	pin control enabled (SSPCON<3:0> =
	0100), the SPI module will reset if the \overline{SS}
	pin is set to VDD.

2: If the SPI is used in Slave mode with CKE set (SSPSTAT register), then the SS pin control must be enabled.

When the SPI module resets, the bit counter is forced to '0'. This can be done by either forcing the \overline{SS} pin to a high level, or clearing the SSPEN bit.

To emulate two-wire communication, the SDO pin can be connected to the SDI pin. When the SPI needs to operate as a receiver, the SDO pin can be configured as an input. This disables transmissions from the SDO. The SDI can always be left as an input (SDI function), since it cannot create a bus conflict.

FIGURE 13-3: SLAVE SYNCHRONIZATION WAVEFORM









13.3.6 SLEEP OPERATION

In Master mode, all module clocks are halted, and the transmission/reception will remain in that state until the device wakes from Sleep. After the device returns to normal mode, the module will continue to transmit/receive data.

In Slave mode, the SPI transmit/receive shift register operates asynchronously to the device. This allows the device to be placed in Sleep mode and data to be shifted into the SPI transmit/receive shift register. When all eight bits have been received, the MSSP interrupt flag bit will be set and, if enabled, will wake the device from Sleep.

13.3.7 EFFECTS OF A RESET

A Reset disables the MSSP module and terminates the current transfer.

13.3.8 BUS MODE COMPATIBILITY

Table 13-1 shows the compatibility between the standard SPI modes and the states of the CKP and CKE control bits.

TABLE 13-1: SPI BUS MODES

Standard SPI Mode	Control Bits State			
Terminology	СКР	CKE		
0,0	0	1		
0,1	0	0		
1,0	1	1		
1,1	1	0		

There is also a SMP bit that controls when the data will be sampled.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other RESETS
INTCON	GIE/GIEH	PEIE/GIEL	T0IE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	0000 000u
PIE1	—	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
PIR1	—	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	-000 0000	0000 0000
SSPBUF	SSPBUF Synchronous Serial Port Receive Buffer/Transmit Register					XXXX XXXX	uuuu uuuu			
SSPCON	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000	0000 0000
SSPSTAT	SMP	CKE	D/A	Р	S	R/W	UA	BF	0000 0000	0000 0000
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	1111 1111	1111 1111
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111	1111 1111

TABLE 13-2: REGISTERS ASSOCIATED WITH SPI OPERATION

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by the MSSP in SPI mode.

Note 1: Bit 6 of PORTA, LATA and TRISA are enabled in ECIO and RCIO Oscillator modes only. In all other oscillator modes, they are disabled and read '0'.

13.4 MSSP I²C Operation

The MSSP module in I²C mode, fully implements all master and slave functions (including general call support) and provides interrupts on Start and Stop bits in hardware, to determine a free bus (Multi-Master mode). The MSSP module implements the standard mode specifications, as well as 7-bit and 10-bit addressing.

Two pins are used for data transfer. These are the RC3/SCK/SCL pin, which is the clock (SCL), and the RC4/SDI/SDA pin, which is the data (SDA). The user must configure these pins as inputs or outputs through the TRISC<4:3> bits.

The MSSP module functions are enabled by setting MSSP Enable bit SSPEN of the SSPCON register.

FIGURE 13-6: MSSP BLOCK DIAGRAM (I²C MODE)



The MSSP module has these six registers for I^2C operation:

- MSSP Control Register 1 (SSPCON)
- MSSP Control Register 2 (SSPCON2)
- MSSP STATUS register (SSPSTAT)
- Serial Receive/Transmit Buffer (SSPBUF)
- MSSP Shift Register (SSPSR) Not directly accessible
- MSSP Address register (SSPADD)
- MSSP Mask register (SSPMSK)

The SSPCON register allows control of the I^2C operation. The SSPM<3:0> mode selection bits (SSPCON register) allow one of the following I^2C modes to be selected:

- I²C Master mode, clock = OSC/4 (SSPADD +1)
- I²C Slave mode (7-bit address)
- I²C Slave mode (10-bit address)
- I²C Slave mode (7-bit address), with Start and Stop bit interrupts enabled
- I²C Slave mode (10-bit address), with Start and Stop bit interrupts enabled
- I²C firmware controlled master operation, slave is idle

Selection of any I^2C mode with the SSPEN bit set, forces the SCL and SDA pins to be open drain, provided these pins are programmed to be inputs by setting the appropriate TRISC bits.

13.4.1 SLAVE MODE

In Slave mode, the SCL and SDA pins must be configured as inputs (TRISC<4:3> set). The MSSP module will override the input state with the output data when required (slave-transmitter).

When an address is matched, or the data transfer after an address match is received, the hardware automatically will generate the Acknowledge (\overline{ACK}) pulse and load the SSPBUF register with the received value currently in the SSPSR register.

If either or both of the following conditions are true, the MSSP module will not give this ACK pulse:

- a) The buffer full bit BF (SSPCON register) was set before the transfer was received.
- b) The overflow bit SSPOV (SSPCON register) was set before the transfer was received.

In this event, the SSPSR register value is not loaded into the SSPBUF, but bit SSPIF of the PIR1 register is set. The BF bit is cleared by reading the SSPBUF register, while bit SSPOV is cleared through software.

The SCL clock input must have a minimum high and low for proper operation. The high and low times of the I^2C specification, as well as the requirement of the MSSP module, are shown in timing parameter #100 and parameter #101.

13.4.1.1 Addressing

Once the MSSP module has been enabled, it waits for a Start condition to occur. Following the Start condition, the eight bits are shifted into the SSPSR register. All incoming bits are sampled with the rising edge of the clock (SCL) line. The value of register SSPSR<7:1> is compared to the value of the SSPADD register. The address is compared on the falling edge of the eighth clock (SCL) pulse. If the addresses match, and the BF and SSPOV bits are clear, the following events occur:

- a) The SSPSR register value is loaded into the SSPBUF register.
- b) The buffer full bit BF is set.
- c) An ACK pulse is generated.
- MSSP interrupt flag bit, SSPIF of the PIR1 register, is set on the falling edge of the ninth SCL pulse (interrupt is generated, if enabled).

In 10-bit address mode, two address bytes need to be received by the slave. The five Most Significant bits (MSb) of the first address byte specify if this is a 10-bit address. The R/\overline{W} bit (SSPSTAT register) must specify a write so the slave device will receive the second address byte. For a 10-bit address, the first byte would equal `1111 0 A9 A8 0', where A9 and A8 are the two MSb's of the address.

The sequence of events for 10-bit addressing is as follows, with steps 7-9 for slave-transmitter:

- 1. Receive first (high) byte of address (bit SSPIF of the PIR1 register and bits BF and UA of the SSPSTAT register are set).
- 2. Update the SSPADD register with second (low) byte of address (clears bit UA and releases the SCL line).
- 3. Read the SSPBUF register (clears bit BF) and clear flag bit SSPIF.
- 4. Receive second (low) byte of address (bits SSPIF, BF, and UA are set).
- 5. Update the SSPADD register with the first (high) byte of address. If match releases SCL line, this will clear bit UA.
- 6. Read the SSPBUF register (clears bit BF) and clear flag bit SSPIF.
- 7. Receive Repeated Start condition.
- 8. Receive first (high) byte of address (bits SSPIF and BF are set).
- 9. Read the SSPBUF register (clears bit BF) and clear flag bit SSPIF.

13.4.1.2 Reception

When the R/W bit of the address byte is clear and an address match occurs, the R/W bit of the SSPSTAT register is cleared. The received address is loaded into the SSPBUF register.

When the address <u>byte</u> overflow condition exists, then no Acknowledge (ACK) pulse is given. An overflow condition is defined as either bit BF (SSPSTAT register) is set, or bit SSPOV (SSPCON register) is set.

An MSSP interrupt is generated for each data transfer byte. Flag bit SSPIF of the PIR1 register must be cleared in software. The SSPSTAT register is used to determine the status of the byte.

13.4.1.3 Transmission

When the $R\overline{W}$ bit of the incoming address byte is set and an address match occurs, the $R\overline{W}$ bit of the SSPSTAT register is set. The received address is loaded into the SSPBUF register. The \overline{ACK} pulse will be sent on the ninth bit and pin RC3/SCK/SCL is held low. The transmit data must be loaded into the SSPBUF register, which also loads the SSPSR register. Then pin RC3/SCK/SCL should be enabled by setting bit CKP (SSPCON register). The master must monitor the SCL pin prior to asserting another clock pulse. The slave devices may be holding off the master by stretching the clock. The eight data bits are shifted out on the falling edge of the SCL input. This ensures that the SDA signal is valid during the SCL high time (Figure 13-8).

An MSSP interrupt is generated for each data transfer byte. The SSPIF bit must be cleared in software and the SSPSTAT register is used to determine the status of the byte. The SSPIF bit is set on the falling edge of the ninth clock pulse.

As a slave-transmitter, the \overline{ACK} pulse from the master-receiver is latched on the rising edge of the ninth SCL input pulse. If the SDA line is high (not \overline{ACK}), then the data transfer is complete. When the \overline{ACK} is latched by the slave, the slave logic is reset and the slave monitors for another occurrence of the Start bit. If the SDA line was low (\overline{ACK}), the transmit data must be loaded into the SSPBUF register, which also loads the SSPSR register. Pin RC3/SCK/SCL should be enabled by setting bit CKP.

PIC16F882/883/884/886/887

FIGURE 13-7: I ² C [™] SLAVE M	IODE WAVEFORMS FOR RECEPTION (7-BIT ADDRESS)
$Receiving Address R/\overline{W} = 0$ $SDA = 1$ $SDA = 1$ $AT \setminus A6 \setminus A5 \setminus A4 \setminus A3 \setminus A2 \setminus A1 \setminus F$ $SCL = 1$ $SSPIF$ BF $SSPOV$	ACK Receiving Data ACK Receiving Data Not ACK r ACK D7 D6 D5 D4 D3 D2 D1 20 D7 D6 25 D4 D3 22 D1 20 /9 1 2 3 4 5 6 7 8 9 1 2 3 4 5 6 7 8 9 1 P1 Bus Master Terminates Transfer Bit SSPOV is set because the SSPBUF register is still full
	ACK is not sent —



13.4.2 GENERAL CALL ADDRESS SUPPORT

The addressing procedure for the I^2C bus is such that, the first byte after the Start condition usually determines which device will be the slave addressed by the master. The exception is the general call address, which can address all devices. When this address is used, all devices should, in theory, respond with an Acknowledge.

The general call address is one of eight addresses reserved for specific purposes by the I^2C protocol. It consists of all 0's with R/W = 0.

The general call address is recognized (enabled) when the General Call Enable (GCEN) bit is set (SSPCON2 register). Following a Start bit detect, eight bits are shifted into the SSPSR and the address is compared against the SSPADD. It is also compared to the general call address and fixed in hardware. If the general call address matches, the SSPSR is transferred to the SSPBUF, the BF bit is set (eighth bit), and on the falling edge of the ninth bit (ACK bit), the SSPIF interrupt flag bit is set.

When the interrupt is serviced, the source for the interrupt can be checked by reading the contents of the SSPBUF. The value can be used to determine if the address was device specific or a general call address.

In 10-bit mode, the SSPADD is required to be updated for the second half of the address to match, and the UA bit is set (SSPSTAT register). If the general call address is sampled when the GCEN bit is set, and while the slave is configured in 10-bit address mode, then the second half of the address is not necessary. The UA bit will not be set, and the slave will begin receiving data after the Acknowledge (Figure 13-9).



FIGURE 13-9: SLAVE MODE GENERAL CALL ADDRESS SEQUENCE (7 OR 10-BIT ADDRESS)

13.4.3 MASTER MODE

Master mode of operation is supported by interrupt generation on the detection of the Start and Stop conditions. The Stop (P) and Start (S) bits are cleared from a Reset, or when the MSSP module is disabled. Control of the I^2C bus may be taken when the P bit is set, or the bus is idle, with both the S and P bits clear.

In Master mode, the SCL and SDA lines are manipulated by the MSSP hardware.

The following events will cause SSP Interrupt Flag bit, SSPIF, to be set (SSP Interrupt if enabled):

- Start condition
- Stop condition
- Data transfer byte transmitted/received
- Acknowledge transmit
- Repeated Start condition

13.4.4 I²C[™] MASTER MODE SUPPORT

Master mode is enabled by setting and clearing the appropriate SSPM bits in SSPCON and by setting the SSPEN bit. Once Master mode is enabled, the user has the following six options:

- 1. Assert a Start condition on SDA and SCL.
- 2. Assert a Repeated Start condition on SDA and SCL.
- 3. Write to the SSPBUF register initiating transmission of data/address.
- 4. Generate a Stop condition on SDA and SCL.
- 5. Configure the I^2C port to receive data.
- 6. Generate an Acknowledge condition at the end of a received byte of data.

Note:	The MSSP module, when configured in I ² C					
	Master mode, does not allow queueing of					
	events. For instance, the user is not					
	allowed to initiate a Start condition and					
	immediately write the SSPBUF register to					
	imitate transmission, before the Start					
	condition is complete. In this case, the					
	SSPBUF will not be written to and the					
	WCOL bit will be set, indicating that a write					
	to the SSPBUF did not occur.					

FIGURE 13-10: MSSP BLOCK DIAGRAM (I²C[™] MASTER MODE)



13.4.4.1 I²C[™] Master Mode Operation

The master device generates all of the serial clock pulses and the Start and Stop conditions. A transfer is ended with a Stop condition or with a Repeated Start condition. Since the Repeated Start condition is also the beginning of the next serial transfer, the I²C bus will not be released.

In Master Transmitter mode, serial data is output through SDA, while SCL outputs the serial clock. The first byte transmitted contains the slave address of the receiving device (7 bits) and the Read/Write (R/W) bit. In this case, the R/W bit will be logic '0'. Serial data is transmitted eight bits at a time. After each byte is transmitted, an Acknowledge bit is received. Start and Stop conditions are output to indicate the beginning and the end of a serial transfer.

In Master Receive mode, the first byte transmitted contains the slave address of the transmitting device (7 bits) and the R/W bit. In this case, the R/W bit will be logic '1'. Thus, the first byte transmitted is a 7-bit slave address followed by a '1' to indicate receive bit. Serial data is received via SDA, while SCL outputs the serial clock. Serial data is received eight bits at a time. After each byte is received, an Acknowledge bit is transmitted. Start and Stop conditions indicate the beginning and end of transmission.

The Baud Rate Generator used for the SPI mode operation is now used to set the SCL clock frequency for either 100 kHz, 400 kHz, or 1 MHz I²C operation. The Baud Rate Generator reload value is contained in the lower 7 bits of the SSPADD register. The Baud Rate Generator will automatically begin counting on a write to the SSPBUF. Once the given operation is complete (i.e., transmission of the last data bit is followed by ACK), the internal clock will automatically stop counting and the SCL pin will remain in its last state. A typical transmit sequence would go as follows:

- a) The user generates a Start condition by setting the Start Enable (SEN) bit (SSPCON2 register).
- SSPIF is set. The MSSP module will wait the required start time before any other operation takes place.
- c) The user loads the SSPBUF with the address to transmit.
- d) Address is shifted out the SDA pin until all eight bits are transmitted.
- e) The MSSP module shifts in the ACK bit from the slave device and writes its value into the ACKSTAT bit (SSPCON2 register).
- f) The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSPIF bit.
- g) The user loads the SSPBUF with eight bits of data.
- h) Data is shifted out the SDA pin until all eight bits are transmitted.
- i) The MSSP module shifts in the ACK bit from the slave device and writes its value into the ACKSTAT bit (SSPCON2 register).
- j) The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSPIF bit.
- k) The user generates a Stop condition by setting the Stop Enable bit PEN (SSPCON2 register).
- I) Interrupt is generated once the Stop condition is complete.

13.4.5 BAUD RATE GENERATOR

In I²C Master mode, the reload value for the BRG is located in the lower 7 bits of the SSPADD register (Figure 13-11). When the BRG is loaded with this value, the BRG counts down to 0 and stops until another reload has taken place. The BRG count is decremented twice per instruction cycle (Tcr) on the Q2 and Q4 clocks. In I²C Master mode, the BRG is reloaded automatically. If clock arbitration is taking place, for instance, the BRG will be reloaded when the SCL pin is sampled high (Figure 13-12).

FIGURE 13-11: BAUD RATE GENERATOR BLOCK DIAGRAM



FIGURE 13-12: BAUD RATE GENERATOR TIMING WITH CLOCK ARBITRATION



13.4.6 I²C[™] MASTER MODE START CONDITION TIMING

To initiate a Start condition, the user sets the Start Condition Enable bit SEN of the SSPCON2 register. If the SDA and SCL pins are sampled high, the Baud Rate Generator is reloaded with the contents of SSPADD<6:0> and starts its count. If SCL and SDA are both sampled high when the Baud Rate Generator times out (TBRG), the SDA pin is driven low. The action of the SDA being driven low, while SCL is high, is the Start condition, and causes the S bit of the SSPSTAT register to be set. Following this, the Baud Rate Generator is reloaded with the contents of SSPADD<6:0> and resumes its count. When the Baud Rate Generator times out (TBRG), the SEN bit of the SSPCON2 register will be automatically cleared by hardware, the Baud Rate Generator is suspended leaving the SDA line held low and the Start condition is complete.

Note: If, at the beginning of the Start condition, the SDA and SCL pins are already sampled low, or if during the Start condition the SCL line is sampled low before the SDA line is driven low, a bus collision occurs, the Bus Collision Interrupt Flag, BCLIF, is set, the Start condition is aborted, and the I²C module is reset into its Idle state.

13.4.6.1 WCOL Status Flag

If the user writes the SSPBUF when a Start sequence is in progress, the WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).

Note: Because queueing of events is not allowed, writing to the lower 5 bits of SSPCON2 is disabled until the Start condition is complete.

FIGURE 13-13: FIRST START BIT TIMING



13.4.7 I²C[™] MASTER MODE REPEATED START CONDITION TIMING

A Repeated Start condition occurs when the RSEN bit (SSPCON2 register) is programmed high and the I²C Logic module is in the Idle state. When the RSEN bit is set, the SCL pin is asserted low. When the SCL pin is sampled low, the Baud Rate Generator is loaded with the contents of SSPADD<5:0> and begins counting. The SDA pin is released (brought high) for one Baud Rate Generator count (TBRG). When the Baud Rate Generator times out, if SDA is sampled high, the SCL pin will be de-asserted (brought high). When SCL is sampled high, the Baud Rate Generator is reloaded with the contents of SSPADD<6:0> and begins counting. SDA and SCL must be sampled high for one TBRG. This action is then followed by assertion of the SDA pin (SDA = 0) for one TBRG, while SCL is high. Following this, the RSEN bit (SSPCON2 register) will be automatically cleared and the Baud Rate Generator will not be reloaded, leaving the SDA pin held low. As soon as a Start condition is detected on the SDA and SCL pins, the S bit (SSPSTAT register) will be set. The SSPIF bit will not be set until the Baud Rate Generator has timed out.



- **2:** A bus collision during the Repeated Start condition occurs if:
 - SDA is sampled low when SCL goes from low-to-high.
 - SCL goes low before SDA is asserted low. This may indicate that another master is attempting to transmit a data "1".

Immediately following the SSPIF bit getting set, the user may write the SSPBUF with the 7-bit address in 7-bit mode, or the default first address in 10-bit mode. After the first eight bits are transmitted and an ACK is received, the user may then transmit an additional eight bits of address (10-bit mode), or eight bits of data (7-bit mode).

13.4.7.1 WCOL Status Flag

If the user writes the SSPBUF when a Repeated Start sequence is in progress, the WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).

Note: Because queueing of events is not allowed, writing of the lower 5 bits of SSPCON2 is disabled until the Repeated Start condition is complete.



FIGURE 13-14: REPEAT START CONDITION WAVEFORM

13.4.8 I²C[™] MASTER MODE TRANSMISSION

Transmission of a data byte, a 7-bit address, or the other half of a 10-bit address, is accomplished by simply writing a value to the SSPBUF register. This action will set the Buffer Full bit, BF, and allow the Baud Rate Generator to begin counting and start the next transmission. Each bit of address/data will be shifted out onto the SDA pin after the falling edge of SCL is asserted (see data hold time specification, parameter 106). SCL is held low for one Baud Rate Generator rollover count (TBRG). Data should be valid before SCL is released high (see data setup time specification, parameter 107). When the SCL pin is released high, it is held that way for TBRG. The data on the SDA pin must remain stable for that duration and some hold time after the next falling edge of SCL. After the eighth bit is shifted out (the falling edge of the eighth clock). the BF bit is cleared and the master releases SDA, allowing the slave device being addressed to respond with an ACK bit during the ninth bit time, if an address match occurs, or if data was received properly. The status of ACK is written into the ACKDT bit on the falling edge of the ninth clock. If the master receives an Acknowledge, the Acknowledge Status bit, ACKSTAT, is cleared. If not, the bit is set. After the ninth clock, the SSPIF bit is set and the master clock (Baud Rate Generator) is suspended until the next data byte is loaded into the SSPBUF, leaving SCL low and SDA unchanged (Figure 13-15).

After the write to the SSPBUF, each bit of the address will be shifted out on the falling edge of SCL, until all seven address bits and the R/W bit, are completed. On the falling edge of the eighth clock, the master will de-assert the SDA pin, allowing the slave to respond with an Acknowledge. On the falling edge of the ninth clock, the master will sample the SDA pin to see if the address was recognized by a slave. The status of the ACK bit is loaded into the ACKSTAT Status bit (SSPCON2 register). Following the falling edge of the ninth clock transmission of the address, the SSPIF is set, the BF bit is cleared and the Baud Rate Generator is turned off, until another write to the SSPBUF takes place, holding SCL low and allowing SDA to float.

13.4.8.1 BF Status Flag

In Transmit mode, the BF bit (SSPSTAT register) is set when the CPU writes to SSPBUF, and is cleared when all eight bits are shifted out.

13.4.8.2 WCOL Status Flag

If the user writes the SSPBUF when a transmit is already in progress (i.e., SSPSR is still shifting out a data byte), the WCOL is set and the contents of the buffer are unchanged (the write doesn't occur). WCOL must be cleared in software.

13.4.8.3 ACKSTAT Status Flag

In Transmit mode, the ACKSTAT bit (SSPCON2 register) is cleared when the slave has sent an Acknowledge ($\overline{ACK} = 0$), and is set when the slave does not Acknowledge ($\overline{ACK} = 1$). A slave sends an Acknowledge when it has recognized its address (including a general call), or when the slave has properly received its data.

13.4.9 I^2C^{TM} MASTER MODE RECEPTION

Master mode reception is enabled by programming the Receive Enable bit, RCEN (SSPCON2 register).

Note:	The MSSP module must be in an Idle state
	before the RCEN bit is set, or the RCEN bit
	will be disregarded.

The Baud Rate Generator begins counting, and on each rollover, the state of the SCL pin changes (high-to-low/low-to-high) and data is shifted into the SSPSR. After the falling edge of the eighth clock, the RCEN bit is automatically cleared, the contents of the SSPSR are loaded into the SSPBUF, the BF bit is set, the SSPIF flag bit is set and the Baud Rate Generator is suspended from counting, holding SCL low. The MSSP is now in Idle state, awaiting the next command. When the buffer is read by the CPU, the BF bit is automatically cleared. The user can then send an Acknowledge bit at the end of reception, by setting the Acknowledge Sequence Enable bit ACKEN (SSPCON2 register).

13.4.9.1 BF Status Flag

In receive operation, the BF bit is set when an address or data byte is loaded into SSPBUF from SSPSR. It is cleared when the SSPBUF register is read.

13.4.9.2 SSPOV Status Flag

In receive operation, the SSPOV bit is set when eight bits are received into the SSPSR and the BF bit is already set from a previous reception.

13.4.9.3 WCOL Status Flag

If the user writes the SSPBUF when a receive is already in progress (i.e., SSPSR is still shifting in a data byte), the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur).

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13.4.10 ACKNOWLEDGE SEQUENCE TIMING

An Acknowledge sequence is enabled by setting the Acknowledge Sequence Enable bit, ACKEN (SSPCON2 register). When this bit is set, the SCL pin is pulled low and the contents of the Acknowledge Data bit (ACKDT) is presented on the SDA pin. If the user wishes to generate an Acknowledge, then the ACKDT bit should be cleared. If not, the user should set the ACKDT bit before starting an Acknowledge sequence. The Baud Rate Generator then counts for one rollover period (TBRG) and the SCL pin is de-asserted (pulled high). When the SCL pin is sampled high (clock arbitration), the Baud Rate Generator counts for TBRG. The SCL pin is then pulled low. Following this, the ACKEN bit is automatically cleared, the Baud Rate Generator is turned off and the MSSP module then goes into Idle mode (Figure 13-17).

13.4.10.1 WCOL Status Flag

If the user writes the SSPBUF when an Acknowledge sequence is in progress, then WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).

13.4.11 STOP CONDITION TIMING

A Stop bit is asserted on the SDA pin at the end of a receive/transmit by setting the Stop Sequence Enable bit, PEN (SSPCON2 register). At the end of a receive/transmit, the SCL line is held low after the falling edge of the ninth clock. When the PEN bit is set, the master will assert the SDA line low. When the SDA line is sampled low, the Baud Rate Generator is reloaded and counts down to 0. When the Baud Rate Generator times out, the SCL pin will be brought high, and one TBRG (Baud Rate Generator rollover count) later, the SDA pin will be de-asserted. When the SDA pin is sampled high while SCL is high, the P bit (SSPSTAT register) is set. A TBRG later, the PEN bit is cleared and the SSPIF bit is set (Figure 13-18).

13.4.11.1 WCOL Status Flag

If the user writes the SSPBUF when a Stop sequence is in progress, then the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur).







13.4.12 CLOCK ARBITRATION

Clock arbitration occurs when the master, during any receive, transmit or Repeated Start/Stop condition, de-asserts the SCL pin (SCL allowed to float high). When the SCL pin is allowed to float high, the Baud Rate Generator (BRG) is suspended from counting until the SCL pin is actually sampled high. When the SCL pin is sampled high, the Baud Rate Generator is reloaded with the contents of SSPADD<6:0> and begins counting. This ensures that the SCL high time will always be at least one BRG rollover count, in the event that the clock is held low by an external device (Figure 13-19).

13.4.13 SLEEP OPERATION

While in Sleep mode, the I²C module can receive addresses or data, and when an address match or complete byte transfer occurs, wake the processor from Sleep (if the MSSP interrupt is enabled).

13.4.14 EFFECT OF A RESET

A Reset disables the MSSP module and terminates the current transfer.

FIGURE 13-19: CLOCK ARBITRATION TIMING IN MASTER TRANSMIT MODE



13.4.15 MULTI-MASTER MODE

In Multi-Master mode, the interrupt generation on the detection of the Start and Stop conditions allows the determination of when the bus is free. The Stop (P) and Start (S) bits are cleared from a Reset, or when the MSSP module is disabled. Control of the I^2C bus may be taken when the P bit (SSPSTAT register) is set, or the bus is idle with both the S and P bits clear. When the bus is busy, enabling the SSP Interrupt will generate the interrupt when the Stop condition occurs.

In Multi-Master operation, the SDA line must be monitored for arbitration, to see if the signal level is the expected output level. This check is performed in hardware, with the result placed in the BCLIF bit.

Arbitration can be lost in the following states:

- Address transfer
- Data transfer
- · A Start condition
- A Repeated Start condition
- An Acknowledge condition

13.4.16 MULTI -MASTER COMMUNICATION, BUS COLLISION, AND BUS ARBITRATION

Multi-Master mode support is achieved by bus arbitration. When the master outputs address/data bits onto the SDA pin, arbitration takes place when the master outputs a '1' on SDA, by letting SDA float high and another master asserts a '0'. When the SCL pin floats high, data should be stable. If the expected data on SDA is a '1' and the data sampled on the SDA pin = 0, then a bus collision has taken place. The master will set the Bus Collision Interrupt Flag (BCLIF) and reset the $I^{2}C$ port to its Idle state (Figure 13-20).

If a transmit was in progress when the bus collision occurred, the transmission is halted, the BF bit is cleared, the SDA and SCL lines are de-asserted, and the SSPBUF can be written to. When the user services the bus collision interrupt service routine, and if the I^2C bus is free, the user can resume communication by asserting a Start condition.

If a Start, Repeated Start, Stop, or Acknowledge condition was in progress when the bus collision occurred, the condition is aborted, the SDA and SCL lines are de-asserted, and the respective control bits in the SSPCON2 register are cleared. When the user services the bus collision interrupt service routine, and if the I^2C bus is free, the user can resume communication by asserting a Start condition.

The master will continue to monitor the SDA and SCL pins. If a Stop condition occurs, the SSPIF bit will be set.

A write to the SSPBUF will start the transmission of data at the first data bit, regardless of where the transmitter left off when the bus collision occurred.

In Multi-Master mode, the interrupt generation on the detection of Start and Stop conditions allows the determination of when the bus is free. Control of the I^2C bus can be taken when the P bit is set in the SSPSTAT register, or the bus is idle and the S and P bits are cleared.

FIGURE 13-20: BUS COLLISION TIMING FOR TRANSMIT AND ACKNOWLEDGE



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13.4.16.1 Bus Collision During a Start Condition

During a Start condition, a bus collision occurs if:

- a) SDA or SCL are sampled low at the beginning of the Start condition (Figure 13-21).
- b) SCL is sampled low before SDA is asserted low (Figure 13-22).

During a Start condition, both the SDA and the SCL pins are monitored, if:

the SDA pin is already low, or the SCL pin is already low,

then:

the Start condition is aborted, and the BCLIF flag is set, and the MSSP module is reset to its Idle state (Figure 13-21).

The Start condition begins with the SDA and SCL pins de-asserted. When the SDA pin is sampled high, the Baud Rate Generator is loaded from SSPADD<6:0> and counts down to 0. If the SCL pin is sampled low

while SDA is high, a bus collision occurs, because it is assumed that another master is attempting to drive a data '1' during the Start condition.

If the SDA pin is sampled low during this count, the BRG is reset and the SDA line is asserted early (Figure 13-23). If, however, a '1' is sampled on the SDA pin, the SDA pin is asserted low at the end of the BRG count. The Baud Rate Generator is then reloaded and counts down to 0, and during this time, if the SCL pin is sampled as '0', a bus collision does not occur. At the end of the BRG count, the SCL pin is asserted low.

Note: The reason that bus collision is not a factor during a Start condition, is that no two bus masters can assert a Start condition at the exact same time. Therefore, one master will always assert SDA before the other. This condition does not cause a bus collision, because the two masters must be allowed to arbitrate the first address following the Start condition. If the address is the same, arbitration must be allowed to continue into the data portion, Repeated Start or Stop conditions.













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13.4.16.2 Bus Collision During a Repeated Start Condition

During a Repeated Start condition, a bus collision occurs if:

- a) A low level is sampled on SDA when SCL goes from low level to high level.
- b) SCL goes low before SDA is asserted low, indicating that another master is attempting to transmit a data '1'.

When the user de-asserts SDA and the pin is allowed to float high, the BRG is loaded with SSPADD<6:0> and counts down to 0. The SCL pin is then de-asserted, and when sampled high, the SDA pin is sampled.

If SDA is low, a bus collision has occurred (i.e, another master is attempting to transmit a data '0', see Figure 13-24). If SDA is sampled high, the BRG is reloaded and begins counting. If SDA goes from high-to-low before the BRG times out, no bus collision occurs because no two masters can assert SDA at exactly the same time.

If SCL goes from high-to-low before the BRG times out and SDA has not already been asserted, a bus collision occurs. In this case, another master is attempting to transmit a data '1' during the Repeated Start condition (Figure 13-25).

If at the end of the BRG time-out, both SCL and SDA are still high, the SDA pin is driven low and the BRG is reloaded and begins counting. At the end of the count, regardless of the status of the SCL pin, the SCL pin is driven low and the Repeated Start condition is complete.

FIGURE 13-24: BUS COLLISION DURING A REPEATED START CONDITION (CASE 1)



FIGURE 13-25: BUS COLLISION DURING REPEATED START CONDITION (CASE 2)



13.4.16.3 Bus Collision During a Stop Condition

Bus collision occurs during a Stop condition if:

- a) After the SDA pin has been de-asserted and allowed to float high, SDA is sampled low after the BRG has timed out.
- b) After the SCL pin is de-asserted, SCL is sampled low before SDA goes high.

The Stop condition begins with SDA asserted low. When SDA is sampled low, the SCL pin is allowed to float. When the pin is sampled high (clock arbitration), the Baud Rate Generator is loaded with SSPADD<6:0> and counts down to 0. After the BRG times out, SDA is sampled. If SDA is sampled low, a bus collision has occurred. This is due to another master attempting to drive a data '0' (Figure 13-26). If the SCL pin is sampled low before SDA is allowed to float high, a bus collision occurs. This is another case of another master attempting to drive a data '0' (Figure 13-27).

FIGURE 13-26: BUS COLLISION DURING A STOP CONDITION (CASE 1)



FIGURE 13-27: BUS COLLISION DURING A STOP CONDITION (CASE 2)



13.4.17 SSP MASK REGISTER

An SSP Mask (SSPMSK) register is available in I^2C Slave mode as a mask for the value held in the SSPSR register during an address comparison operation. A zero ('0') bit in the SSPMSK register has the effect of making the corresponding bit in the SSPSR register a "don't care".

This register is reset to all '1's upon any Reset condition and, therefore, has no effect on standard SSP operation until written with a mask value.

This register must be initiated prior to setting SSPM<3:0> bits to select the I^2C Slave mode (7-bit or 10-bit address).

This register can only be accessed when the appropriate mode is selected by bits (SSPM<3:0> of SSPCON).

The SSP Mask register is active during:

- - - - - -

- 7-bit Address mode: address compare of A<7:1>.
- 10-bit Address mode: address compare of A<7:0> only. The SSP mask has no effect during the reception of the first (high) byte of the address.

- . . .

- . . .

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	
MSK7	MSK6	MSK5	MSK4	MSK3	MSK2	MSK1	MSK0 ⁽²⁾	
bit 7				-			bit 0	
Legend:								
R = Readable bit W = Writable bit				U = Unimplemented bit, read as '0'				
-n = Value at P	OR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown	
bit 7-1	MSK<7:1>: N	lask bits						

- . . .

REGISTER 13-4: SSPMSK: SSP MASK REGISTER⁽¹⁾

	 1 = The received address bit n is compared to SSPADD<n> to detect I²C address match</n> 0 = The received address bit n is not used to detect I²C address match
bit 0	MSK<0>: Mask bit for I ² C Slave mode, 10-bit Address ⁽²⁾
	I ² C Slave mode, 10-bit Address (SSPM<3:0> = 0111):
	1 = The received address bit 0 is compared to SSPADD<0> to detect I ² C address match
	0 = The received address bit 0 is not used to detect I ² C address match

Note 1: When SSPCON bits SSPM<3:0> = 1001, any reads or writes to the SSPADD SFR address are accessed through the SSPMSK register.

2: In all other SSP modes, this bit has no effect.

14.0 SPECIAL FEATURES OF THE CPU

The PIC16F882/883/884/886/887 have a host of features intended to maximize system reliability, minimize cost through elimination of external components, provide power-saving features and offer code protection.

These features are:

- Reset
 - Power-on Reset (POR)
 - Power-up Timer (PWRT)
 - Oscillator Start-up Timer (OST)
 - Brown-out Reset (BOR)
- Interrupts
- Watchdog Timer (WDT)
- Oscillator selection
- Sleep
- Code protection
- ID Locations
- In-Circuit Serial Programming™

The PIC16F882/883/884/886/887 have two timers that offer necessary delays on power-up. One is the Oscillator Start-up Timer (OST), intended to keep the chip in Reset until the crystal oscillator is stable. The other is the Power-up Timer (PWRT), which provides a fixed delay of 64 ms (nominal) on power-up only, designed to keep the part in Reset while the power supply stabilizes. There is also circuitry to reset the device if a brown-out occurs, which can use the Powerup Timer to provide at least a 64 ms Reset. With these three functions-on-chip, most applications need no external Reset circuitry.

The Sleep mode is designed to offer a very low-current Power-Down mode. The user can wake-up from Sleep through:

- External Reset
- Watchdog Timer Wake-up
- An interrupt

Several oscillator options are also made available to allow the part to fit the application. The INTOSC option saves system cost while the LP crystal option saves power. A set of Configuration bits are used to select various options (see Register 14-3).

14.1 Configuration Bits

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The Configuration bits can be programmed (read as '0'), or left unprogrammed (read as '1') to select various device configurations as shown in Register 14-1. These bits are mapped in program memory location 2007h.

Note: Address 2007h is beyond the user program memory space. It belongs to the special configuration memory space (2000h-3FFFh), which can be accessed only during programming. See "*PIC16F88X Memory Programming Specification*" (DS41287) for more information.

REGISTER 14-1: CONFIG1: CONFIGURATION WORD REGISTER 1

—	—	DEBUG	LVP	FCMEN	IESO	BOREN1	BOREN0
bit 15							bit 8

CPD	CP	MCLRE	PWRTE	WDTE	FOSC2	FOSC1	FOSC0
bit 7							bit 0

bit 15-14	Unimplemented: Read as '1'
bit 13	DEBUG: In-Circuit Debugger Mode bit 1 = In-Circuit Debugger disabled, RB6/ICSPCLK and RB7/ICSPDAT are general purpose I/O pins 0 = In-Circuit Debugger enabled, RB6/ICSPCLK and RB7/ICSPDAT are dedicated to the debugger
bit 12	LVP: Low Voltage Programming Enable bit 1 = RB3/PGM pin has PGM function, low voltage programming enabled 0 = RB3 pin is digital I/O, HV on MCLR must be used for programming
bit 11	FCMEN: Fail-Safe Clock Monitor Enabled bit 1 = Fail-Safe Clock Monitor is enabled 0 = Fail-Safe Clock Monitor is disabled
bit 10	IESO: Internal External Switchover bit 1 = Internal/External Switchover mode is enabled 0 = Internal/External Switchover mode is disabled
bit 9-8	BOREN<1:0>: Brown-out Reset Selection bits ⁽¹⁾ 11 = BOR enabled 10 = BOR enabled during operation and disabled in Sleep 01 = BOR controlled by SBOREN bit of the PCON register 00 = BOR disabled
bit 7	CPD: Data Code Protection bit ⁽²⁾ 1 = Data memory code protection is disabled 0 = Data memory code protection is enabled
bit 6	CP: Code Protection bit ⁽³⁾ 1 = Program memory code protection is disabled 0 = Program memory code protection is enabled
bit 5	MCLRE: RE3/MCLR pin function select bit ⁽⁴⁾ 1 = RE3/ <u>MCLR</u> pin function is MCLR 0 = RE3/MCLR pin function is digital input, MCLR internally tied to VDD
bit 4	PWRTE: Power-up Timer Enable bit 1 = PWRT disabled 0 = PWRT enabled
bit 3	WDTE: Watchdog Timer Enable bit 1 = WDT enabled 0 = WDT disabled and can be enabled by SWDTEN bit of the WDTCON register
bit 2-0	FOSC<2:0>: Oscillator Selection bits 111 = RC oscillator: CLKOUT function on RA6/OSC2/CLKOUT pin, RC on RA7/OSC1/CLKIN 100 = RCIO oscillator: I/O function on RA6/OSC2/CLKOUT pin, RC on RA7/OSC1/CLKIN 101 = INTOSC oscillator: CLKOUT function on RA6/OSC2/CLKOUT pin, I/O function on RA7/OSC1/CLKIN 100 = INTOSCIO oscillator: I/O function on RA6/OSC2/CLKOUT pin, I/O function on RA7/OSC1/CLKIN 101 = EC: I/O function on RA6/OSC2/CLKOUT pin, CLKIN on RA7/OSC1/CLKIN 102 = HS oscillator: High-speed crystal/resonator on RA6/OSC2/CLKOUT and RA7/OSC1/CLKIN 103 = XT oscillator: Crystal/resonator on RA6/OSC2/CLKOUT and RA7/OSC1/CLKIN 104 = LP oscillator: Low-power crystal on RA6/OSC2/CLKOUT and RA7/OSC1/CLKIN
Note 1: 2:	Enabling Brown-out Reset does not automatically enable Power-up Timer. The entire data EEPROM will be erased when the code protection is turned off.

- 3: The entire program memory will be erased when the code protection is turned off.
- 4: When MCLR is asserted in INTOSC or RC mode, the internal clock oscillator is disabled.

REGISTER 14-2: CONFIG2: CONFIGURATION WORD REGISTER 2

_	—	—	—	—	WRT1	WRT0	BOR4V
15							bit 8
15				1			_

—	—	—	—	—	—	—	—
bit 7							bit 0

bit 15-11	Unimplemented: Read as '1'						
bit 10-9	WRT<1:0>: Flash Program Memory Self Write Enable bits						
	PIC16F883/PIC16F884						
	00 = 0000h to 07FFh write protected, 0800h to 0FFFh may be modified by EECON control						
	01 = 0000h to 03FFh write protected, 0400h to 0FFFh may be modified by EECON control						
	10 = 0000h to 00FFh write protected, 0100h to 0FFFh may be modified by EECON control						
	11 = Write protection off						
	PIC16F886/PIC16F887						
	00 = 0000h to 0FFFh write protected, 1000h to 1FFFh may be modified by EECON control						
	01 = 0000h to 07FFh write protected, 0800h to 1FFFh may be modified by EECON control						
	10 = 0000h to 00FFh write protected, 0100h to 1FFFh may be modified by EECON control						
	11 = Write protection off						
	<u>PIC16F882</u>						
	00 = 0000h to 03FFh write protected, 0400h to 07FFh may be modified by EECON control						
	01 = 0000h to 00FFh write protected, 0100h to 07FFh may be modified by EECON control						
	11 = Write protection off						
bit 8	BOR4V: Brown-out Reset Selection bit						
	0 = Brown-out Reset set to 2.1V						

- 1 = Brown-out Reset set to 4.0V
- bit 7-0 Unimplemented: Read as '1'

14.2 Reset

The PIC16F882/883/884/886/887 differentiates between various kinds of Reset:

- a) Power-on Reset (POR)
- b) WDT Reset during normal operation
- c) WDT Reset during Sleep
- d) MCLR Reset during normal operation
- e) MCLR Reset during Sleep
- f) Brown-out Reset (BOR)

Some registers are not affected in any Reset condition; their status is unknown on POR and unchanged in any other Reset. Most other registers are reset to a "Reset state" on:

- Power-on Reset
- MCLR Reset
- MCLR Reset during Sleep
- WDT Reset
- Brown-out Reset (BOR)

They are not affected by a WDT Wake-up since this is viewed as the resumption of normal operation. TO and PD bits are set or cleared differently in different Reset situations, as indicated in Table 14-2. These bits are used in software to determine the nature of the Reset. See Table 14-5 for a full description of Reset states of all registers.

A simplified block diagram of the On-Chip Reset Circuit is shown in Figure 14-1.

The MCLR Reset path has a noise filter to detect and ignore small pulses. See **Section 17.0** "**Electrical Specifications**" for pulse-width specifications.

FIGURE 14-1: SIMPLIFIED BLOCK DIAGRAM OF ON-CHIP RESET CIRCUIT



14.2.1 POWER-ON RESET (POR)

The on-chip POR circuit holds the chip in Reset until VDD has reached a high enough level for proper operation. A maximum rise time for VDD is required. See **Section 17.0 "Electrical Specifications"** for details. If the BOR is enabled, the maximum rise time specification does not apply. The BOR circuitry will keep the device in Reset until VDD reaches VBOR (see **Section 14.2.4** "**Brown-out Reset (BOR)**").

Note:	The POR circuit does not produce an
	internal Reset when VDD declines. To
	re-enable the POR, VDD must reach Vss
	for a minimum of 100 μs.

When the device starts normal operation (exits the Reset condition), device operating parameters (i.e., voltage, frequency, temperature, etc.) must be met to ensure operation. If these conditions are not met, the device must be held in Reset until the operating conditions are met.

For additional information, refer to Application Note AN607, *"Power-up Trouble Shooting"* (DS00607).

14.2.2 MCLR

 $\frac{\text{PIC16F882}/\text{883}/\text{884}/\text{886}/\text{887}}{\text{MCLR}} \text{ Reset path. The filter will detect and ignore small pulses.}$

It should be noted that a WDT Reset does not drive MCLR pin low.

The behavior of the ESD protection on the MCLR pin has been altered from early devices of this family. Voltages applied to the pin that exceed its specification can result in both MCLR Resets and excessive current beyond the device specification during the ESD event. For this reason, Microchip recommends that the MCLR pin no longer be tied directly to VDD. The use of an RC network, as shown in Figure 14-2, is suggested.

An internal $\overline{\text{MCLR}}$ option is enabled by clearing the $\overline{\text{MCLRE}}$ bit in the Configuration Word Register 1. When $\overline{\text{MCLRE}} = 0$, the Reset signal to the chip is generated internally. When the $\overline{\text{MCLRE}} = 1$, the RA3/ $\overline{\text{MCLR}}$ pin becomes an external Reset input. In this mode, the RA3/ $\overline{\text{MCLR}}$ pin has a weak pull-up to VDD.

FIGURE 14-2: RECOMMENDED MCLR CIRCUIT



14.2.3 POWER-UP TIMER (PWRT)

The Power-up Timer provides a fixed 64 ms (nominal) time-out on power-up only, from POR or Brown-out Reset. The Power-up Timer operates from the 31 kHz LFINTOSC oscillator. For more information, see **Section 4.5 "Internal Clock Modes"**. The chip is kept in Reset as long as PWRT is active. The PWRT delay allows the VDD to rise to an acceptable level. A Configuration bit, PWRTE, can disable (if set) or enable (if cleared or programmed) the Power-up Timer. The Power-up Timer should be enabled when Brown-out Reset is enabled, although it is not required.

The Power-up Timer delay will vary from chip-to-chip and vary due to:

- VDD variation
- Temperature variation
- Process variation

See DC parameters for details (Section 17.0 "Electrical Specifications").

14.2.4 BROWN-OUT RESET (BOR)

The BOREN0 and BOREN1 bits in the Configuration Word Register 1 select one of four BOR modes. Two modes have been added to allow software or hardware control of the BOR enable. When BOREN<1:0> = 01, the SBOREN bit (PCON<4>) enables/disables the BOR allowing it to be controlled in software. By selecting BOREN<1:0>, the BOR is automatically disabled in Sleep to conserve power and enabled on wake-up. In this mode, the SBOREN bit is disabled. See Register 14-3 for the Configuration Word definition.

The BOR4V bit in the Configuration Word Register 2 selects one of two Brown-out Reset voltages. When BOR4B = 1, VBOR is set to 4V. When BOR4V = 0, VBOR is set to 2.1V.

If VDD falls below VBOR for greater than parameter (TBOR) (see **Section 17.0 "Electrical Specifications"**), the Brown-out situation will reset the device. This will

occur regardless of VDD slew rate. A Reset is not insured to occur if VDD falls below VBOR for less than parameter (TBOR).

On any Reset (Power-on, Brown-out Reset, Watchdog Timer, etc.), the chip will remain in Reset until VDD rises above VBOR (see Figure 14-3). The Power-up Timer will now be invoked, if enabled and will keep the chip in Reset an additional 64 ms.

Note:	The Power-up Timer is enabled by the
	PWRTE bit in the Configuration Word
	Register 1.

If VDD drops below VBOR while the Power-up Timer is running, the chip will go back into a Brown-out Reset and the Power-up Timer will be re-initialized. Once VDD rises above VBOR, the Power-up Timer will execute a 64 ms Reset.



FIGURE 14-3: BROWN-OUT SITUATIONS

14.2.5 TIME-OUT SEQUENCE

On power-up, the time-out sequence is as follows: first, PWRT time-out is invoked after POR has expired, then OST is activated after the PWRT time-out has expired. The total time-out will vary based on oscillator configuration and <u>PWRTE</u> bit status. For example, in EC mode with <u>PWRTE</u> bit erased (PWRT disabled), there will be no time-out at all. Figures 14-4, 14-5 and 14-6 depict time-out sequences. The device can execute code from the INTOSC while OST is active by enabling Two-Speed Start-up or Fail-Safe Monitor (see Section 4.7.2 "Two-speed Start-up Sequence" and Section 4.8 "Fail-Safe Clock Monitor").

Since the time-outs occur from the POR pulse, if MCLR is kept low long enough, the time-outs will expire. Then, bringing MCLR high will begin execution immediately (see Figure 14-5). This is useful for testing purposes or to synchronize more than one PIC16F882/883/884/ 886/887 device operating in parallel.

Table 14-5 shows the Reset conditions for some special registers, while Table 14-4 shows the Reset conditions for all the registers.

14.2.6 POWER CONTROL (PCON) REGISTER

The Power Control register PCON (address 8Eh) has two Status bits to indicate what type of Reset that last occurred.

Bit 0 is $\overline{\text{BOR}}$ (Brown-out Reset). $\overline{\text{BOR}}$ is unknown on Power-on Reset. It must then be set by the user and checked on subsequent Resets to see if $\overline{\text{BOR}} = 0$, indicating that a Brown-out has occurred. The $\overline{\text{BOR}}$ Status bit is a "don't care" and is not necessarily predictable if the brown-out circuit is disabled (BOREN<1:0> = 00 in the Configuration Word Register 1).

Bit 1 is POR (Power-on Reset). It is a '0' on Power-on Reset and unaffected otherwise. The user must write a '1' to this bit following a Power-on Reset. On a subsequent Reset, if POR is '0', it will indicate that a Power-on Reset has occurred (i.e., VDD may have gone too low).

For more information, see Section 3.2.2 "Ultra Low-Power Wake-up" and Section 14.2.4 "Brown-out Reset (BOR)".

Oscillator Configuration	Powe	er-up	Brown-o	Wake-up from		
	PWRTE = 0	PWRTE = 1	PWRTE = 0	PWRTE = 1	Sleep	
XT, HS, LP	TPWRT + 1024 • Tosc	1024 • Tosc	Tpwrt + 1024 • Tosc	1024 • Tosc	1024 • Tosc	
LP, T1OSCIN = 1	TPWRT	—	TPWRT	—	—	
RC, EC, INTOSC	TPWRT		TPWRT	_	—	

TABLE 14-1: TIME-OUT IN VARIOUS SITUATIONS

TABLE 14-2: STATUS/PCON BITS AND THEIR SIGNIFICANCE

POR	BOR	то	PD	Condition			
0	x	1	1	Power-on Reset			
u	0	1	1	Brown-out Reset			
u	u	0	u	WDT Reset			
u	u	0	0	WDT Wake-up			
u	u	u	u	MCLR Reset during normal operation			
u	u	1	0	MCLR Reset during Sleep			

Legend: u = unchanged, x = unknown

TABLE 14-3: SUMMARY OF REGISTERS ASSOCIATED WITH BROWN-OUT

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
PCON	_	_	ULPWUE	SBOREN	_	_	POR	BOR	01qq	0uuu
STATUS	IRP	RP1	RPO	TO	PD	Z	DC	С	0001 1xxx	000q quuu

Legend: u = unchanged, x = unknown, - = unimplemented bit, reads as '0', q = value depends on condition. Shaded cells are not used by BOR. Note 1: Other (non Power-up) Resets include MCLR Reset and Watchdog Timer Reset during normal operation.

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FIGURE 14-4: TIME-OUT SEQUENCE ON POWER-UP (DELAYED MCLR): CASE 1

FIGURE 14-5: TIME-OUT SEQUENCE ON POWER-UP (DELAYED MCLR): CASE 2



FIGURE 14-6: TIME-OUT SEQUENCE ON POWER-UP (MCLR WITH VDD)



Register	Address	Power-on Reset	MCLR Reset WDT Reset Brown-out Reset ⁽¹⁾	Wake-up from Sleep through Interrupt Wake-up from Sleep through WDT Time-out
W	—	xxxx xxxx	uuuu uuuu	uuuu uuuu
INDF	00h/80h/ 100h/180h	XXXX XXXX	XXXX XXXX	սսսս սսսս
TMR0	01h/101h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PCL	02h/82h/ 102h/182h	0000 0000	0000 0000	PC + 1 ⁽³⁾
STATUS	03h/83h/ 103h/183h	0001 1xxx	000q quuu ⁽⁴⁾	uuuq quuu ⁽⁴⁾
FSR	04h/84h/ 104h/184h	XXXX XXXX	սսսս սսսս	սսսս սսսս
PORTA	05h	xxxx xxxx	0000 0000	uuuu uuuu
PORTB	06h/106h	xxxx xxxx	0000 0000	uuuu uuuu
PORTC	07h	xxxx xxxx	0000 0000	uuuu uuuu
PORTD	08h	xxxx xxxx	0000 0000	uuuu uuuu
PORTE	09h	xxxx	0000	uuuu
PCLATH	0Ah/8Ah/ 10Ah/18Ah	0 0000	0 0000	u uuuu
INTCON	0Bh/8Bh/ 10Bh/18Bh	0000 000x	0000 000u	uuuu uuuu ⁽²⁾
PIR1	0Ch	0000 0000	0000 0000	uuuu uuuu (2)
PIR2	0Dh	0000 0000	0000 0000	uuuu uuuu (2)
TMR1L	0Eh	xxxx xxxx	uuuu uuuu	uuuu uuuu
TMR1H	0Fh	xxxx xxxx	uuuu uuuu	uuuu uuuu
T1CON	10h	0000 0000	uuuu uuuu	-uuu uuuu
TMR2	11h	0000 0000	0000 0000	uuuu uuuu
T2CON	12h	-000 0000	-000 0000	-uuu uuuu
SSPBUF	13h	xxxx xxxx	uuuu uuuu	uuuu uuuu
SSPCON	14h	0000 0000	0000 0000	uuuu uuuu
CCPR1L	15h	xxxx xxxx	uuuu uuuu	uuuu uuuu
CCPR1H	16h	xxxx xxxx	<u>uuuu</u> uuuu	นนนน นนนน
CCP1CON	17h	0000 0000	0000 0000	<u>uuuu</u> <u>uuuu</u>
RCSTA	18h	0000 000x	0000 0000	<u>uuuu</u> uuuu
TXREG	19h	0000 0000	0000 0000	<u>uuuu</u> <u>uuuu</u>
RCREG	1Ah	0000 0000	0000 0000	uuuu uuuu
CCPR2L	1Bh	xxxx xxxx	uuuu uuuu	uuuu uuuu

TABLE 14-4: INITIALIZATION CONDITION FOR REGISTER

Legend: u = unchanged, x = unknown, - = unimplemented bit, reads as '0', q = value depends on condition.

Note 1: If VDD goes too low, Power-on Reset will be activated and registers will be affected differently.

2: One or more bits in INTCON and/or PIR1 will be affected (to cause wake-up).

- **3:** When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).
- 4: See Table 14-5 for Reset value for specific condition.
- **5:** If Reset was due to brown-out, then bit 0 = 0. All other Resets will cause bit 0 = u.
- **6**: Accessible only when SSPCON register bits SSPM<3:0> = 1001.

TABLE 14-4:	INITIALIZATION CONDITION FOR REGISTER (CONTINUED))
-------------	---	---

Register	Address	Power-on Reset	MCLR Reset WDT Reset (Continued) Brown-out Reset ⁽¹⁾	Wake-up from Sleep through Interrupt Wake-up from Sleep through WDT Time-out (Continued)
CCPR2H	1Ch	xxxx xxxx	սսսս սսսս	uuuu uuuu
CCP2CON	1Dh	00 0000	00 0000	uu uuuu
ADRESH	1Eh	xxxx xxxx	սսսս սսսս	սսսս սսսս
ADCON0	1Fh	00-0 0000	00-0 0000	uu-u uuuu
OPTION_REG	81h/181h	1111 1111	1111 1111	uuuu uuuu
TRISA	85h	1111 1111	1111 1111	uuuu uuuu
TRISB	86h/186h	1111 1111	1111 1111	uuuu uuuu
TRISC	87h	1111 1111	1111 1111	uuuu uuuu
TRISD	88h	1111 1111	1111 1111	uuuu uuuu
TRISE	89h	1111	1111	uuuu
PIE1	8Ch	0000 0000	0000 0000	uuuu uuuu
PIE2	8Dh	0000 0000	0000 0000	uuuu uuuu
PCON	8Eh	010x	0uuu (1, 5)	uuuu
OSCCON	8Fh	-110 q000	-110 q000	-uuu uuuu
OSCTUNE	90h	0 0000	u uuuu	u uuuu
SSPCON2	91h	0000 0000	0000 0000	uuuu uuuu
PR2	92h	1111 1111	1111 1111	1111 1111
SSPADD ⁽⁶⁾	93h	0000 0000	0000 0000	uuuu uuuu
SSPMSK ⁽⁶⁾	93h	1111 1111	1111 1111	1111 1111
SSPSTAT	94h	0000 0000	0000 0000	uuuu uuuu
WPUB	95h	1111 1111	1111 1111	uuuu uuuu
IOCB	96h	0000 0000	0000 0000	uuuu uuuu
VRCON	97h	0000 0000	0000 0000	uuuu uuuu
TXSTA	98h	0000 -010	0000 -010	uuuu -uuu
SPBRG	99h	0000 0000	0000 0000	uuuu uuuu
SPBRGH	9Ah	0000 0000	0000 0000	uuuu uuuu
PWM1CON	9Bh	0000 0000	0000 0000	uuuu uuuu
ECCPAS	9Ch	0000 0000	0000 0000	uuuu uuuu
PSTRCON	9Dh	0 0001	0 0001	u uuuu
ADRESL	9Eh	xxxx xxxx	<u>uuuu</u> uuuu	นนนน นนนน
ADCON1	9Fh	0-00	0-00	u-uu
WDTCON	105h	0 1000	0 1000	u uuuu
CM1CON0	107h	0000 0-00	0000 0-00	uuuu u-uu
CM2CON0	108h	0000 0-00	0000 0-00	uuuu u-uu

Legend: u = unchanged, x = unknown, - = unimplemented bit, reads as '0', q = value depends on condition.

Note 1: If VDD goes too low, Power-on Reset will be activated and registers will be affected differently.

2: One or more bits in INTCON and/or PIR1 will be affected (to cause wake-up).

3: When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).

4: See Table 14-5 for Reset value for specific condition.

5: If Reset was due to brown-out, then bit 0 = 0. All other Resets will cause bit 0 = u.

6: Accessible only when SSPCON register bits SSPM<3:0> = 1001.
Register	Address	s Power-on Reset WDT Reset (Continued) Brown-out Reset ⁽¹⁾		Wake-up from Sleep through Interrupt Wake-up from Sleep through WDT Time-out (Continued)
CM2CON1	109h	0000 00	0000 00	uuuu uu
EEDAT	10Ch	0000 0000	0000 0000	uuuu uuuu
EEADR	10Dh	0000 0000	0000 0000	uuuu uuuu
EEDATH	10Eh	00 0000	00 0000	uu uuuu
EEADRH	10Fh	0 0000	0 0000	u uuuu
SRCON	185h	0000 00-0	0000 00-0	uuuu uu-u
BAUDCTL	187h	01-0 0-00	01-0 0-00	uu-u u-uu
ANSEL	188h	1111 1111	1111 1111	uuuu uuuu
ANSELH	189h	1111 1111	1111 1111	uuuu uuuu
EECON1	18Ch	x000	q000	uuuu
EECON2	18Dh			

TABLE 14-4: INITIALIZATION CONDITION FOR REGISTER (CONTINUED)

Legend: u = unchanged, x = unknown, - = unimplemented bit, reads as '0', q = value depends on condition.

Note 1: If VDD goes too low, Power-on Reset will be activated and registers will be affected differently.

- 2: One or more bits in INTCON and/or PIR1 will be affected (to cause wake-up).
- **3:** When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).
- 4: See Table 14-5 for Reset value for specific condition.
- **5:** If Reset was due to brown-out, then bit 0 = 0. All other Resets will cause bit 0 = u.
- 6: Accessible only when SSPCON register bits SSPM<3:0> = 1001.

TABLE 14-5: INITIALIZATION CONDITION FOR SPECIAL REGISTERS

Condition	Program Counter	Status Register	PCON Register
Power-on Reset	000h	0001 1xxx	010x
MCLR Reset during normal operation	000h	000u uuuu	0uuu
MCLR Reset during Sleep	000h	0001 0uuu	0uuu
WDT Reset	000h	0000 uuuu	0uuu
WDT Wake-up	PC + 1	սսս0 Օսսս	uuuu
Brown-out Reset	000h	0001 luuu	01u0
Interrupt Wake-up from Sleep	PC + 1 ⁽¹⁾	uuul 0uuu	uuuu

Legend: u = unchanged, x = unknown, - = unimplemented bit, reads as '0'.

Note 1: When the wake-up is due to an interrupt and Global Interrupt Enable bit, GIE, is set, the PC is loaded with the interrupt vector (0004h) after execution of PC + 1.

14.3 Interrupts

The PIC16F882/883/884/886/887 has multiple interrupt sources:

- External Interrupt RB0/INT
- Timer0 Overflow Interrupt
- PORTB Change Interrupts
- 2 Comparator Interrupts
- A/D Interrupt
- Timer1 Overflow Interrupt
- Timer2 Match Interrupt
- EEPROM Data Write Interrupt
- Fail-Safe Clock Monitor Interrupt
- Enhanced CCP Interrupt
- EUSART Receive and Transmit Interrupts
- Ultra Low-Power Wake-up Interrupt
- MSSP Interrupt

The Interrupt Control register (INTCON) and Peripheral Interrupt Request Register 1 (PIR1) record individual interrupt requests in flag bits. The INTCON register also has individual and global interrupt enable bits.

A Global Interrupt Enable bit, GIE (INTCON<7>), enables (if set) all unmasked interrupts, or disables (if cleared) all interrupts. Individual interrupts can be disabled through their corresponding enable bits in the INTCON, PIE1 and PIE2 registers, respectively. GIE is cleared on Reset.

The Return from Interrupt instruction, RETFIE, exits the interrupt routine, as well as sets the GIE bit, which re-enables unmasked interrupts.

The following interrupt flags are contained in the INTCON register:

- INT Pin Interrupt
- PORTB Change Interrupts
- Timer0 Overflow Interrupt

The peripheral interrupt flags are contained in the PIR1 and PIR2 registers. The corresponding interrupt enable bits are contained in PIE1 and PIE2 registers.

The following interrupt flags are contained in the PIR1 register:

- A/D Interrupt
- EUSART Receive and Transmit Interrupts
- Timer1 Overflow Interrupt
- Synchronous Serial Port (SSP) Interrupt
- Enhanced CCP1 Interrupt
- Timer1 Overflow Interrupt
- Timer2 Match Interrupt

The following interrupt flags are contained in the PIR2 register:

- Fail-Safe Clock Monitor Interrupt
- 2 Comparator Interrupts
- EEPROM Data Write Interrupt
- Ultra Low-Power Wake-up Interrupt
- CCP2 Interrupt

When an interrupt is serviced:

- The GIE is cleared to disable any further interrupt.
- The return address is pushed onto the stack.
- The PC is loaded with 0004h.

For external interrupt events, such as the INT pin, PORTB change interrupts, the interrupt latency will be three or four instruction cycles. The exact latency depends upon when the interrupt event occurs (see Figure 14-8). The latency is the same for one or twocycle instructions. Once in the Interrupt Service Routine, the source(s) of the interrupt can be determined by polling the interrupt flag bits. The interrupt flag bit(s) must be cleared in software before re-enabling interrupts to avoid multiple interrupt requests.

- Note 1: Individual interrupt flag bits are set, regardless of the status of their corresponding mask bit or the GIE bit.
 - 2: When an instruction that clears the GIE bit is executed, any interrupts that were pending for execution in the next cycle are ignored. The interrupts, which were ignored, are still pending to be serviced when the GIE bit is set again.

For additional information on Timer1, Timer2, comparators, A/D, data EEPROM, EUSART, MSSP or Enhanced CCP modules, refer to the respective peripheral section.

14.3.1 RB0/INT INTERRUPT

External interrupt on RB0/INT pin is edge-triggered; either rising if the INTEDG bit (OPTION REG<6>) is set, or falling, if the INTEDG bit is clear. When a valid edge appears on the RB0/INT pin, the INTF bit (INTCON<1>) is set. This interrupt can be disabled by clearing the INTE control bit (INTCON<4>). The INTF bit must be cleared in software in the Interrupt Service Routine before re-enabling this interrupt. The RB0/INT interrupt can wake-up the processor from Sleep, if the INTE bit was set prior to going into Sleep. The status of the GIE bit decides whether or not the processor branches to the interrupt vector following wake-up (0004h). See Section 14.6 "Power-Down Mode (Sleep)" for details on Sleep and Figure 14-10 for timing of wake-up from Sleep through RB0/INT interrupt.

14.3.2 TIMER0 INTERRUPT

An overflow (FFh \rightarrow 00h) in the TMR0 register will set the T0IF (INTCON<2>) bit. The interrupt can be enabled/disabled by setting/clearing T0IE (INTCON<5>) bit. See **Section 5.0 "Timer0 Module"** for operation of the Timer0 module.

14.3.3 PORTB INTERRUPT

An input change on PORTB change sets the RBIF (INTCON<0>) bit. The interrupt can be enabled/ disabled by setting/clearing the RBIE (INTCON<3>) bit. Plus, individual pins can be configured through the IOCB register.

Note: If a change on the I/O pin should occur when the read operation is being executed (start of the Q2 cycle), then the RBIF interrupt flag may not get set. See Section 3.4.3 "Interrupt-on-Change" for more information.





FIGURE 14-8: INT PIN INTERRUPT TIMING

Note 1: INTF flag is sampled here (every Q1).

- 2: Asynchronous interrupt latency = 3-4 TCY. Synchronous latency = 3 TCY, where TCY = instruction cycle time. Latency is the same whether Inst (PC) is a single cycle or a 2-cycle instruction.
- 3: CLKOUT is available only in INTOSC and RC Oscillator modes.
- 4: For minimum width of INT pulse, refer to AC specifications in Section 17.0 "Electrical Specifications".
- 5: INTF is enabled to be set any time during the Q4-Q1 cycles.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
INTCON	GIE	PEIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	0000 000x
PIE1	_	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	-000 0000	-000 0000
PIE2	OSFIE	C2IE	C1IE	EEIE	BCLIE	ULPWUIE	_	CCP2IE	0000 00-0	0000 00-0
PIR1	_	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	-000 0000	-000 0000
PIR2	OSFIF	C2IF	C1IF	EEIF	BCLIF	ULPWUIF	_	CCP2IF	0000 00-0	0000 00-0

TABLE 14-6: SUMMARY OF INTERRUPT REGISTERS

Legend: x = unknown, u = unchanged, - = unimplemented read as '0', q = value depends upon condition. Shaded cells are not used by the Interrupt module.

14.4 Context Saving During Interrupts

During an interrupt, only the return PC value is saved on the stack. Typically, users may wish to save key registers during an interrupt (e.g., W and STATUS registers). This must be implemented in software.

Since the upper 16 bytes of all GPR banks are common in the PIC16F882/883/884/886/887 (see Figures 2-2 and 2-3), temporary holding registers, W_TEMP and STATUS_TEMP, should be placed in here. These 16 locations do not require banking and therefore, make it easier to context save and restore. The same code shown in Example 14-1 can be used to:

- Store the W register
- Store the STATUS register
- Execute the ISR code
- Restore the Status (and Bank Select Bit register)
- Restore the W register

Note: The PIC16F882/883/884/886/887 normally does not require saving the PCLATH. However, if computed GOTO's are used in the ISR and the main code, the PCLATH must be saved and restored in the ISR.

EXAMPLE 14-1: SAVING STATUS AND W REGISTERS IN RAM

```
MOVWF W TEMP
                        ;Copy W to TEMP register
SWAPF STATUS,W
                       ;Swap status to be saved into W
                        ;Swaps are used because they do not affect the status bits
MOVWF STATUS TEMP
                       ;Save status to bank zero STATUS TEMP register
•
:(ISR)
                       ;Insert user code here
SWAPF STATUS TEMP,W
                       ;Swap STATUS TEMP register into W
                        ; (sets bank to original state)
MOVWF STATUS
                       ;Move W into STATUS register
SWAPF W TEMP,F
                       ;Swap W TEMP
SWAPF W_TEMP,W
                       ;Swap W_TEMP into W
```

14.5 Watchdog Timer (WDT)

The WDT has the following features:

- Operates from the LFINTOSC (31 kHz)
- · Contains a 16-bit prescaler
- Shares an 8-bit prescaler with Timer0
- Time-out period is from 1 ms to 268 seconds
- Configuration bit and software controlled

WDT is cleared under certain conditions described in Table 14-7.

14.5.1 WDT OSCILLATOR

The WDT derives its time base from the 31 kHz LFINTOSC. The LTS bit of the OSCCON register does not reflect that the LFINTOSC is enabled.

The value of WDTCON is '---0 1000' on all Resets. This gives a nominal time base of 17 ms.

Note: When the Oscillator Start-up Timer (OST) is invoked, the WDT is held in Reset, because the WDT Ripple Counter is used by the OST to perform the oscillator delay count. When the OST count has expired, the WDT will begin counting (if enabled).

14.5.2 WDT CONTROL

The WDTE bit is located in the Configuration Word Register 1. When set, the WDT runs continuously.

When the WDTE bit in the Configuration Word Register 1 is set, the SWDTEN bit of the WDTCON register has no effect. If WDTE is clear, then the SWDTEN bit can be used to enable and disable the WDT. Setting the bit will enable it and clearing the bit will disable it.

The PSA and PS<2:0> bits of the OPTION register have the same function as in previous versions of the PIC16F882/883/884/886/887 Family of microcontrollers. See **Section 5.0 "Timer0 Module"** for more information.



TABLE 14-7: WDT STATUS

Conditions	WDT
WDTE = 0	Cleared
CLRWDT Command	
Oscillator Fail Detected	
Exit Sleep + System Clock = T1OSC, EXTRC, INTOSC, EXTCLK	
Exit Sleep + System Clock = XT, HS, LP	Cleared until the end of OST

11.0	11.0		DAM 0		D/M/ O	D/M/ C	
0-0	0-0	0-0			R/VV-U	K/W-U	
	—		WDTPS3	WDTPS2	WDTPS1	WDTPS0	SWDTEN
bit 7							bit 0
Legend:							
R = Readable	e bit	W = Writable	bit	U = Unimpler	mented bit, read	l as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unk	nown
bit 7-5	Unimplemen	ted: Read as '	0'				
bit 4-1	WDTPS-3.05	• Watchdog Ti	mer Period Se	lect hits			
	Rit Value = P	rescale Rate					
	0000 = 1.32	roodalo rato					
	0000 = 1:62 0001 = 1:64						
	0010 = 1:12	8					
	0011 = 1:25	6					
	0100 = 1:51	2 (Reset value)				
	0101 = 1:102	24					
	0110 = 1:204	48					
	1000 - 1.819	90 92					
	1000 = 1.013 1001 = 1.163	384					
	1010 = 1:32	768					
	1011 = 1:65	536					
	1100 = rese	rved					
	1101 = rese	rved					
	1110 = rese	rved					
					(1)		
bit 0	SWDTEN: So	oftware Enable	or Disable the	Watchdog Tir	ner		
	1 = WDT is tu	Irned on	(
	0 = WDT is tu	irned off (Rese	t value)				
Note 1: If	WDTE configur onfiguration bit =	ration bit = 1, = 0, then it is p	then WDT is ossible to turn	always enable WDT on/off w	ed, irrespective ith this control b	of this contro bit.	I bit. If WDTE

REGISTER 14-3: WDTCON: WATCHDOG TIMER CONTROL REGISTER

 TABLE 14-8:
 SUMMARY OF WATCHDOG TIMER REGISTER

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets	
CONFIG1 ⁽¹⁾	CPD	CP	MCLRE	PWRTE	WDTE	FOSC2	FOSC1	FOSC0	_		
OPTION_REG	RBPU	INTEDG	TOCS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111	
WDTCON		—	_	WDTPS3	WDTPS2	WSTPS1	WDTPS0	SWDTEN	0 1000	0 1000	

Legend: Shaded cells are not used by the Watchdog Timer.

Note 1: See Register 14-1 for operation of all Configuration Word Register 1 bits.

14.6 Power-Down Mode (Sleep)

The Power-down mode is entered by executing a $\ensuremath{\mathtt{SLEEP}}$ instruction.

If the Watchdog Timer is enabled:

- WDT will be cleared but keeps running.
- PD bit in the STATUS register is cleared.
- TO bit is set.
- Oscillator driver is turned off.
- I/O ports maintain the status they had before SLEEP was executed (driving high, low or highimpedance).

For lowest current consumption in this mode, all I/O pins should be either at VDD or VSS, with no external circuitry drawing current from the I/O pin and the comparators and CVREF should be disabled. I/O pins that are highimpedance inputs should be pulled high or low externally to avoid switching currents caused by floating inputs. The T0CKI input should also be at VDD or VSS for lowest current consumption. The contribution from on-chip pullups on PORTA should be considered.

The $\overline{\text{MCLR}}$ pin must be at a logic high level.

Note: It should be noted that a Reset generated by a WDT time-out does not drive MCLR pin low.

14.6.1 WAKE-UP FROM SLEEP

The device can wake-up from Sleep through one of the following events:

- 1. External Reset input on $\overline{\text{MCLR}}$ pin.
- 2. Watchdog Timer Wake-up (if WDT was enabled).
- 3. Interrupt from RB0/INT pin, PORTB change or a peripheral interrupt.

The first event will cause a device Reset. The two latter events are considered a continuation of program execution. The \overline{TO} and \overline{PD} bits in the STATUS register can be used to determine the cause of device Reset. The \overline{PD} bit, which is set on power-up, is cleared when Sleep is invoked. \overline{TO} bit is cleared if WDT Wake-up occurred.

The following peripheral interrupts can wake the device from Sleep:

- 1. TMR1 interrupt. Timer1 must be operating as an asynchronous counter.
- 2. ECCP Capture mode interrupt.
- 3. A/D conversion (when A/D clock source is FRC).
- 4. EEPROM write operation completion.
- 5. Comparator output changes state.
- 6. Interrupt-on-change.
- 7. External Interrupt from INT pin.
- 8. EUSART Break detect, I²C slave.

Other peripherals cannot generate interrupts since during Sleep, no on-chip clocks are present.

When the SLEEP instruction is being executed, the next instruction (PC + 1) is prefetched. For the device to wake-up through an interrupt event, the corresponding interrupt enable bit must be set (enabled). Wake-up occurs regardless of the state of the GIE bit. If the GIE bit is clear (disabled), the device continues execution at the instruction after the SLEEP instruction. If the GIE bit is set (enabled), the device executes the instruction after the SLEEP instruction, then branches to the interrupt address (0004h). In cases where the execution of the instruction following SLEEP is not desirable, the user should have a NOP after the SLEEP instruction.

Note: If the global interrupts are disabled (GIE is cleared), but any interrupt source has both its interrupt enable bit and the corresponding interrupt flag bits set, the device will immediately wake-up from Sleep. The SLEEP instruction is completely executed.

The WDT is cleared when the device wakes up from Sleep, regardless of the source of wake-up.

14.6.2 WAKE-UP USING INTERRUPTS

When global interrupts are disabled (GIE cleared) and any interrupt source has both its interrupt enable bit and interrupt flag bit set, one of the following will occur:

- If the interrupt occurs **before** the execution of a SLEEP instruction, the SLEEP instruction will complete as a NOP. Therefore, the WDT and WDT prescaler and postscaler (if enabled) will not be cleared, the TO bit will not be set and the PD bit will not be cleared.
- If the interrupt occurs during or after the execution of a SLEEP instruction, the device will immediately wake-up from Sleep. The SLEEP instruction will be completely executed before the wake-up. Therefore, the WDT and WDT prescaler and postscaler (if enabled) will be cleared, the TO bit will be set and the PD bit will be cleared.

Even if the flag bits were checked before executing a SLEEP instruction, it may be possible for flag bits to become set before the SLEEP instruction completes. To determine whether a SLEEP instruction executed, test the PD bit. If the PD bit is set, the SLEEP instruction was executed as a NOP.

To ensure that the WDT is cleared, a CLRWDT instruction should be executed before a SLEEP instruction.

FIGURE 14-10:	WAKE-	UP FROM	I SLEEP	THROU	JGH INTER	RUPT		
	22 Q3 Q4; Q1 \/	Q2 Q3 Q4; /_/_/\/ /\	Q1	Tost ⁽²⁾	01 Q2 Q3 Q4 \	Q1 Q2 Q3 Q4;	Q1 Q2 Q3 Q4; (Q1 Q2 Q3 Q4; \
INTE flag	I			1		· · ·		i
(INTCON<1>)					Interrupt Laten	су ⁽³⁾		
GIE bit (INTCON<7>)			Processor ir Sleep			<u>. </u>	'-	
Instruction Flow	1	:		1		1 i	1	1
PC X	PC X	PC + 1	C PC ·	+2 X	PC + 2	X PC + 2	(<u>0004h</u>)	0005h
Instruction { Inst(I Fetched { Inst(I	PC) = Sleep I	nst(PC + 1)		1	Inst(PC + 2)	 	Inst(0004h)	Inst(0005h)
Instruction { In Executed { In	st(PC – 1)	Sleep		1	Inst(PC + 1)	Dummy Cycle	Dummy Cycle	Inst(0004h)
Note 1: XT, H 2: Tost 3: GIF -	S or LP Oscillato = 1024 Tosc (dra	or mode assum awing not to so	ned. cale). This del	ay does no	ot apply to EC an	nd RC Oscillator m	iodes.	line
.				F. 5000000	,			

4: CLKOUT is not available in XT, HS, LP or EC Oscillator modes, but shown here for timing reference.

14.7 Code Protection

If the code protection bit(s) have not been programmed, the on-chip program memory can be read out using ICSP[™] for verification purposes.

Note:	The entire data EEPROM and Flash
	program memory will be erased when the
	code protection is switched from on to off.
	See the "PIC16F88X Memory Programming
	Specification" (DS41287) for more
	information.

14.8 ID Locations

Four memory locations (2000h-2003h) are designated as ID locations where the user can store checksum or other code identification numbers. These locations are not accessible during normal execution but are readable and writable during Program/Verify mode. Only the Least Significant 7 bits of the ID locations are used.

14.9 In-Circuit Serial Programming™

The PIC16F882/883/884/886/887 microcontrollers can be serially programmed while in the end application circuit. This is simply done with two lines for clock and data and three other lines for:

- power
- ground
- programming voltage

This allows customers to manufacture boards with unprogrammed devices and then program the microcontroller just before shipping the product. This also allows the most recent firmware or a custom firmware to be programmed. The device is placed into a Program/Verify mode by holding the RB6/ICSPCLK and RB7/ICSPDAT pins low, while raising the MCLR (VPP) pin from VIL to VIHH. See the "*PIC16F88X Memory Programming Specification*" (DS41287) for more information. RB7 becomes the programming data and RB0 becomes the programming clock. Both RB7 and RB0 are Schmitt Trigger inputs in this mode.

After Reset, to place the device into Program/Verify mode, the Program Counter (PC) is at location 00h. A 6-bit command is then supplied to the device. Depending on the command, 14 bits of program data are then supplied to or from the device, depending on whether the command was a Load or a Read. For complete details of serial programming, please refer to the "*PIC16F88X Memory Programming Specification*" (DS41287).

A typical In-Circuit Serial Programming connection is shown in Figure 14-11.

FIGURE 14-11:

TYPICAL IN-CIRCUIT SERIAL PROGRAMMING™ CONNECTION



14.10 In-Circuit Debugger

The PIC16F882/883/884/886/887-ICD can be used in any of the package types. The device will be mounted on the target application board, which in turn has a 3 or 4 wire connection to the ICD tool.

When the debug bit in the Configuration Word (CONFIG<13>) is programmed to a '0', the In-Circuit Debugger functionality is enabled. This function allows simple debugging functions when used with MPLAB[®] ICD 2. When the microcontroller has this feature enabled, some of the resources are not available for general use. See Table 14-9 for more detail.

Note: The user's application must have the circuitry required to support ICD functionality. Once the ICD circuitry is enabled, normal device pin functions on RB6/ICSPCLK and RB7/ICSPDAT will not be usable. The ICD circuitry uses these pins for communication with the ICD2 external debugger.

For more information, see "Using *MPLAB*[®] *ICD* 2" (DS51265), available on Microchip's web site (www.microchip.com).

14.10.1 ICD PINOUT

The devices in the PIC16F88X family carry the circuitry for the In-Circuit Debugger on-chip and on existing device pins. This eliminates the need for a separate die or package for the ICD device. The pinout for the ICD device is the same as the devices (see **Section 1.0 "Device Overview"** for complete pinout and pin descriptions). Table 14-9 shows the location and function of the ICD related pins on the 28 and 40 pin devices.

TABLE 14-9:	PIC16F883/884/886/887-ICD PIN DESCRIPTIONS

Pin (PDIP)								
PIC16F884/887	PIC16F882/883/ 886	Name	Туре	Pull-up	Description			
40	28	ICDDATA	TTL	_	In-Circuit Debugger Bidirectional data			
39	27	ICDCLK	ST		In-Circuit Debugger Bidirectional clock			
1	1	MCLR/VPP	ΗV	—	Programming voltage			
11,32	20	VDD	Р					
12,31	8,19	Vss	Р					

Legend: TTL = TTL input buffer, ST = Schmitt Trigger input buffer, P = Power, HV = High Voltage

15.0 INSTRUCTION SET SUMMARY

The PIC16F883/884/886/887 instruction set is highly orthogonal and is comprised of three basic categories:

- Byte-oriented operations
- Bit-oriented operations
- Literal and control operations

Each PIC16 instruction is a 14-bit word divided into an **opcode**, which specifies the instruction type and one or more **operands**, which further specify the operation of the instruction. The formats for each of the categories is presented in Figure 15-1, while the various opcode fields are summarized in Table 15-1.

Table 15-2 lists the instructions recognized by the MPASMTM assembler.

For **byte-oriented** instructions, 'f' represents a file register designator and 'd' represents a destination designator. The file register designator specifies which file register is to be used by the instruction.

The destination designator specifies where the result of the operation is to be placed. If 'd' is zero, the result is placed in the W register. If 'd' is one, the result is placed in the file register specified in the instruction.

For **bit-oriented** instructions, 'b' represents a bit field designator, which selects the bit affected by the operation, while 'f' represents the address of the file in which the bit is located.

For **literal and control** operations, 'k' represents an 8-bit or 11-bit constant, or literal value.

One instruction cycle consists of four oscillator periods; for an oscillator frequency of 4 MHz, this gives a normal instruction execution time of 1 μ s. All instructions are executed within a single instruction cycle, unless a conditional test is true, or the program counter is changed as a result of an instruction. When this occurs, the execution takes two instruction cycles, with the second cycle executed as a NOP.

All instruction examples use the format '0xhh' to represent a hexadecimal number, where 'h' signifies a hexadecimal digit.

15.1 Read-Modify-Write Operations

Any instruction that specifies a file register as part of the instruction performs a Read-Modify-Write (RMW) operation. The register is read, the data is modified, and the result is stored according to either the instruction, or the destination designator 'd'. A read operation is performed on a register even if the instruction writes to that register.

For example, a CLRF PORTA instruction will read PORTA, clear all the data bits, then write the result back to PORTA. This example would have the unintended consequence of clearing the condition that set the RAIF flag.

TABLE 15-1: OPCODE FIELD DESCRIPTIONS

Field	Description
f	Register file address (0x00 to 0x7F)
W	Working register (accumulator)
b	Bit address within an 8-bit file register
k	Literal field, constant data or label
x	Don't care location (= 0 or 1). The assembler will generate code with x = 0 . It is the recommended form of use for compatibility with all Microchip software tools.
d	Destination select; $d = 0$: store result in W, d = 1: store result in file register f. Default is $d = 1$.
PC	Program Counter
TO	Time-out bit
С	Carry bit
DC	Digit carry bit
Z	Zero bit
PD	Power-down bit

FIGURE 15-1: GENERAL FORMAT FOR INSTRUCTIONS



Operands Description Cycles MSb LSb Affected Notes BYTE-ORIENTED FILE REGISTER OPERATIONS ADDWF f, d Add W and f 1 00 0111 dfff ffff Z 1, 2 ANDWW f, d AND W with f 1 00 0011 dff ffff Z 2 1, 2 CLRF f Clear M 1 00 0010 dff ffff Z 1, 2	Mnemo	onic,				14-Bit	Opcode	9	Status	
BYTE-ORIENTED FILE REGISTER OPERATIONS ADDWF f, d Add W and f 1 00 0111 dfff ffff Z 1, 2 CLRF f Clear f 1 00 0001 1fff ffff Z 2 1, 2 CLRV - Clear W 1 00 0010 dfff ffff Z 2 COMF f, d Decrement f 1 00 0010 dffff Ffff Z 1, 2, 3 DECFSZ f, d Increment f, Skip if 0 1(2) 00 1011 dffffff Z 1, 2, 3 INCF f, d Increment f, Skip if 0 1(2) 00 11 dfffffff 1, 2, 3 INCFF f, d Move f 1 00 1000 dfffffff 1, 2, 3 MOVF f, d Rotate Left through Carry 1 00 1000 dfff 1, 2 SUBAPF f, d Subtract Wrom f 1 00	Opera	nds	Description	Cycles	MSb			LSb	Affected	Notes
ADDWF f, d Add W and f 1 00 0111 dfff ffff Z 1, 2 ANDWF f, d AND W with f 1 00 0101 dfff ffff Z 1, 2 CLRF f Clear f 1 00 0001 dfff ffff Z 1, 2 COMF f, d Decrement f 1 00 0010 dfff fff Z 1, 2 DECF f, d Decrement f, Skip if 0 11 00 1010 dfff ffff Z 1, 2, 3 INCF f, d Increment f, Skip if 0 12(2) 00 1111 dfff fff Z 1, 2 3 INCF f, d Increment f, Skip if 0 1 100 1000 dfff fff Z 1, 2 3 INCF f, d Increment f, Skip if Carry 1 00 1000 dfff ffff Z 1, 2 1, 2 SUBWF <th></th> <th></th> <th>BYTE-ORIENTED FILE REG</th> <th>STER OPE</th> <th>RATIO</th> <th>NS</th> <th></th> <th></th> <th></th> <th></th>			BYTE-ORIENTED FILE REG	STER OPE	RATIO	NS				
ANDWF f,d AND W with f 1 00 0101 dfff fff Z 1.2 CLRF f Clear W 1 00 0001 lfff fff Z 2 COMF f,d Complement f 1 00 0001 dfff fff Z 1,2 DECF f,d Decrement f, Skip if 0 1(2) 00 0101 dfff fff Z 1,2,3 INCF f,d Increment f, Skip if 0 1(2) 00 1010 dfff fff Z 1,2,3 INCFS f,d Increment f, Skip if 0 1(2) 00 1111 dfff fff Z 1,2,3 INCFS f,d Increment f, Skip if 0 1(2) 00 1100 dfff fff Z 1,2,2 MOVF f,d Move f 1 00 0000 0x00 0x00 0x00 0x00 0x00 0x00 1,2 1,2,2 1,2	ADDWF	f, d	Add W and f	1	00	0111	dfff	ffff	C, DC, Z	1, 2
CLRF f Clear f 1 00 0001 lfff fff Z 2 CCMF f, d Complement f 1 00 0001 dfff fff Z 1 2 2 DECF f, d Decrement f 1 00 1011 dfff fff Z 1, 2 1, 2, 3 DECFSZ f, d Increment f, Skip if 0 1(2) 00 1011 dfff fff Z 1, 2, 3 INCF f, d Increment f, Skip if 0 1(2) 00 1111 dfff T, 2 1, 2 1, 2 MOVF f, d Move OW with f 1 00 1000 dfff ffff Z 1, 2 MOVF f, d Rotate Left through Carry 1 00 1000 dfff ffff C 1, 2 SUBWF f, d Subtract W from f 1 00 1000 dfff ffff Z 1, 2 SUBWF	ANDWF	f, d	AND W with f	1	00	0101	dfff	ffff	Z	1, 2
CLRW - Clear W 1 00 0001 0xx xxxx Z COMF f, d Complement f 1 00 1001 dfff fff Z 1, 2 DECF f, d Decrement f, Skip if 0 1(2) 00 1011 dfff fff Z 1, 2 INCF f, d Increment f, Skip if 0 1(2) 00 1011 dfff fff Z 1, 2 INCF f, d Inclusive OR W with f 1 00 1010 dfff fff Z 1, 2 INCF f, d Inclusive OR W with f 1 00 1000 dfff fff Z 1, 2 INCF f, d Inclusive OR W with f 1 00 1000 dfff fff Z 1, 2 MOVF f Move f Move f 1 00 1000 0000 0xxx wxx Z 1, 2 SUBWF f, d Subtract W from f 1	CLRF	f	Clear f	1	00	0001	lfff	ffff	Z	2
COMF f, d Complement f 1 00 1001 dfff ffff Z 1, 2 DECFS f, d Decrement f, Skip if 0 1(2) 00 1011 dfff ffff 1, 2 INCF f, d Increment f, Skip if 0 1(2) 00 1010 dfff fff 1, 2, 3 INCFSZ f, d Increment f, Skip if 0 1(2) 00 1010 dfff fff Z 1, 2 INCFSZ f, d Move f 1 00 1000 dfff ffff Z 1, 2 MOVF f, d Move W to f 1 00 1000 dfff ffff Z 1, 2 NOP - No Operation 1 00 1000 dfff ffff C 1, 2 SUBWF f, d Rotate Right fthrough Carry 1 00 1000 dfff ffff C, DC, Z 1, 2 SUBWF f, d Subtract W from f 1	CLRW	-	Clear W	1	00	0001	0xxx	xxxx	Z	
DECF f, d Decrement f 1 0 0 0.011 dfff ffff Z 1, 2 DECFSZ f, d Increment fspi f0 1(2) 00 1011 dfff ffff Z 1, 2 3 INCF f, d Increment f 1 00 1000 dfff ffff Z 1, 2 INCF f, d Increment f 1 00 1000 dfff ffff Z 1, 2 INCF f, d Increment f 1 00 1000 dfff ffff Z 1, 2 MOVF f Move W to f 1 00 000 dfff ffff Z 1, 2 MOVF f Move M to f 1 00 0100 dfff fff Z 1, 2 SUBWF f, d Rotate Edift through Carry 1 00 1100 dfff fff Z 1, 2 SUBWF f, d Subtract W from f<	COMF	f, d	Complement f	1	00	1001	dfff	ffff	Z	1, 2
DECFSZ f, d Decrement f, Skip if 0 1(2) 00 1011 dfff ffff 1, 2, 3 INCF f, d Increment f 1 00 1010 dfff ffff Z 1, 2, 3 INCFSZ f, d Increment f, Skip if 0 1(2) 00 1010 dfff ffff Z 1, 2, 3 INCFSZ f, d Increment f, Skip if 0 1(2) 00 1010 dfff ffff Z 1, 2, 3 MOVF f, d Move f 1 00 0000 dfff ffff Z 1, 2 MOVF f, d Move f 1 00 0000 dff ffff Z 1, 2 MOVF f, d Rotate Right fthrough Carry 1 00 1100 dfff C, DC, Z 1, 2 SUBWF f, d Subtract W from f 1 00 1100 dfff ffff Z 1, 2 SUBWF f, d Subtract W from f 1	DECF	f, d	Decrement f	1	00	0011	dfff	ffff	Z	1, 2
INCF f, d Increment f 1 00 1010 dfff fff Z 1, 2 INCFSZ f, d Increment f, Skip if 0 1(2) 00 1111 dff ffff Z 1, 2 1, 2 1, 2, 3 IORWF f, d Inclusive OR W with f 1 00 1010 dfff ffff Z 1, 2 MOVF f, d Move W to f 1 00 1000 dfff ffff Z 1, 2 MOVF f Move W to f 1 00 1000 dfff ffff Z 1, 2 MOVF f, d Rotate Leff through Carry 1 00 1100 dfff ffff C 1, 2 SUBWF f, d Subtract W from f 1 00 1100 dfff ffff Z 1, 2 XORWF f, d Exclusive OR W with f 1 00 1010 dfff ffff Z 1, 2 SWAPF f, d Exclusive OR W with f 1 01 010 bb fff fff	DECFSZ	f, d	Decrement f, Skip if 0	1 (2)	00	1011	dfff	ffff		1, 2, 3
INCFSZ f, d Increment f, Skip if 0 1(2) 00 1111 dfff ffff 1, 2, 3 IORWF f, d Inclusive OR W with f 1 00 0100 dfff ffff Z 1, 2 MOVF f Move f 1 00 0100 dfff ffff Z 1, 2 MOVF f Move W to f 1 00 0000 0xx0 0000 RLF f, d Rotate Left through Carry 1 00 1010 dfff ffff C 1, 2 SUBWF f, d Subtract W from f 1 00 0100 dfff ffff C 1, 2 SWAPF f, d Swap nibbles in f 1 00 0110 dfff ffff Z 1, 2 SWEF f, d Bit Clear f 1 1 01 000bb bfff ffff 1, 2 SWAPF f, b Bit Clear f 1 1 1 1 1, 2 BTFS f, b Bit Test f, Skip if Clear	INCF	f, d	Increment f	1	00	1010	dfff	ffff	Z	1, 2
IORWF f, d Inclusive OR W with f 1 00 0100 dfff ffff Z 1, 2 MOVF f, d Move f 1 00 0000 dfff ffff Z 1, 2 MOVWF f Move W to f 1 00 0000 dfff ffff Z 1, 2 MOVWF f Move W to f 1 00 0000 0xx0 0000 C 1, 2 RRF f, d Rotate Left fthrough Carry 1 00 1010 dfff ffff C 1, 2 SUBWF f, d Subtract W from f 1 00 0110 dfff ffff C 1, 2 1, 2 SUBWF f, d Subtract W from f 1 00 0110 dfff ffff C D, D	INCFSZ	f, d	Increment f, Skip if 0	1 (2)	00	1111	dfff	ffff		1, 2, 3
MOVF f, d Move f Move f 1 00 1000 dfff ffff Z 1, 2 MOVF f Move W to f 1 0 0000 1fff ffff C 1, 2 NOP - No Operation 1 0 0000 0xx0 0000 1101 dfff ffff C 1, 2 RRF f, d Rotate Left through Carry 1 00 1100 dfff ffff C 1, 2 SUBWF f, d Subtract W from f 1 0 110 dfff ffff C 1, 2 XORWF f, d Swap nibbles in f 1 00 1100 dfff ffff Z 1, 2 XORWF f, d Swap nibbles in f 1 01 00bb bfff ffff Z 1, 2 SUBF f, b Bit Clear f 1 1 01 01bb bfff ffff 3 3 DTFSS	IORWF	f, d	Inclusive OR W with f	1	00	0100	dfff	ffff	Z	1, 2
MOVWF f Move W to f 1 00 00000 lfff ffff NOP - No Operation 1 00 0000 0xxx0 0000 RLF f, d Rotate Left through Carry 1 00 1101 dfff ffff C 1, 2 SUBWF f, d Subtract W from f 1 00 1100 dfff ffff C, DC, Z 1, 2 SWAPF f, d Swap nibbles in f 1 00 0110 dfff ffff Z 1, 2 XORWF f, d Exclusive OR W with f 1 00 0110 dfff ffff Z 1, 2 XORWF f, d Exclusive OR W with f 1 01 00bb bfff ffff Z 1, 2 SWAPF f, d Bit Clear f 1 1 01 00bb bfff ffff 1, 2 BSF f, b Bit Test f, Skip if Clear 1 1 11 111	MOVF	f, d	Move f	1	00	1000	dfff	ffff	Z	1, 2
NOP - No Operation 1 00 0000 0xxx 0000 RLF f, d Rotate Left fthrough Carry 1 00 1101 dfff fffff C 1, 2 RRF f, d Subtract W from f 1 00 1100 dfff ffff C 1, 2 SUBWF f, d Swap nibbles in f 1 00 0110 dfff ffff C, DC, Z 1, 2 XORWF f, d Exclusive OR W with f 1 00 0110 dfff ffff Z 1, 2 XORWF f, d Exclusive OR W with f 1 01 00bb bfff fffff Z 1, 2 XORWF f, d Bit Clear f 1 01 01bb bfff fffff Z 1, 2 BSF f, b Bit Set f 1 01 10bb bfff fffff 3 3 BTFSS f, b Bit Test f, Skip if Clear 1 1 11 111x kkkk kkkk Z Z ADDLW k Ad	MOVWF	f	Move W to f	1	00	0000	lfff	ffff		
RLF f, d Rotate Left f through Carry 1 00 1101 dfff ff C 1, 2 RRF f, d Subtract W from f 1 00 1100 dfff ffff C 1, 2 SUBWF f, d Swap nibbles in f 1 00 1100 dfff ffff C, DC, Z 1, 2 SWAPF f, d Exclusive OR W with f 1 00 1100 dfff ffff Z, Z 1, 2 XORWF f, d Exclusive OR W with f 1 01 010 bb bfff ffff Z 1, 2 BCF f, b Bit Clear f 1 01 010 bb bfff ffff 3 BTFSC f, b Bit Test f, Skip if Clear 1 1 11 1100 bb/ff ffff 3 BTFSS f, b Bit Test f, Skip if Set 1 1 11 1100 kkkkkkk Z ADDLW k Add literal and W	NOP	-	No Operation	1	00	0000	0xx0	0000		
RRF f, d Rotate Right f through Carry 1 00 1100 dfff ffff C 1, 2 SUBWF f, d Swap nibbles in f 1 00 010 dfff ffff C, DC, Z 1, 2 XORWF f, d Exclusive OR W with f 1 00 0110 dfff ffff Z, 2 1, 2 XORWF f, d Exclusive OR W with f 1 00 0110 dfff ffff Z 1, 2 XORWF f, d Exclusive OR W with f 1 01 00bb bfff ffff Z 1, 2 BCF f, b Bit Clear f 1 1 01 00bb bfff ffff 1, 2 BTFSC f, b Bit Test f, Skip if Clear 1 1 1 1 1 1 1 1 1 2 3 DLW k Add literal and W 1 1 1 1 1 1 1 1	RLF	f, d	Rotate Left f through Carry	1	00	1101	dfff	ffff	С	1, 2
SUBWF f, d Subtract W from f 1 00 0010 dfff ffff C, DC, Z 1, 2 SWAPF f, d Swap nibbles in f 1 00 0110 dfff ffff Z 1, 2 XORWF f, d Exclusive OR W with f 1 00 0110 dfff ffff Z 1, 2 BIT-ORIENTED FILE REGISTER OPERATIONS BCF f, b Bit Clear f 1 01 00bb bfff ffff 1, 2 BIT-ORIENTED FILE REGISTER OPERATIONS BCF f, b Bit Clear f 1 01 00bb bfff ffff 1, 2 BTFSC f, b Bit Test f, Skip if Clear 1 (2) 01 10bb bfff ffff 3 LITERAL AND CONTROL OPERATIONS ADDLW k Add literal and W 1 11 1111x kkkk Kkkk Z CALL k Call Subroutine 2 10 0kkkk kkk	RRF	f, d	Rotate Right f through Carry	1	00	1100	dfff	ffff	С	1, 2
SWAPF XORWF f, d Swap nibbles in f Exclusive OR W with f 1 00 1110 dfff ffff 1, 2 BIT-ORIENTED FILE REGISTER OPERATIONS BCF f, b Bit Clear f 1 01 00bb bfff ffff 1, 2 1, 2 BIT-ORIENTED FILE REGISTER OPERATIONS BCF f, b Bit Clear f 1 01 00bb bfff ffff 1, 2 1, 2 BTFSC f, b Bit Test f, Skip if Clear 1 01 01bb bfff ffff 3 LITERAL AND CONTROL OPERATIONS Add literal and W ANDLW k Add literal and W 1 11 111 111x kkkk kkkk Z Z ANDLW k Call Subroutine 2 10 0kkk kkkk Z Z Z Z Z Z Z Z Z D Dkkk kkkk Z Z Z Z Z Z D Dkkk kk	SUBWF	f, d	Subtract W from f	1	00	0010	dfff	ffff	C, DC, Z	1, 2
XORWF f, d Exclusive OR W with f 1 00 0110 dfff fff Z 1, 2 BIT-ORIENTED FILE REGISTER OPERATIONS BCF f, b Bit Clear f 1 01 00bb bfff ffff 1, 2 BSF f, b Bit Set f 1 01 01bb bfff ffff 1, 2 BTFSS f, b Bit Test f, Skip if Clear 1 1 01 01bb bfff ffff 3 BTFSS f, b Bit Test f, Skip if Set 1 1 11 11bb bfff ffff 3 ADDLW k Add literal and W 1 1 11 1010 kkkk kkkk Z Z ANDLW k Add literal with W 1 1 11 1001 kkkk kkkk Z Z CALL k Call Subroutine 2 10 0kkk kkkk Z Z TO, PD TO, PD TO, PD GO to a	SWAPF	f, d	Swap nibbles in f	1	00	1110	dfff	ffff		1, 2
BIT-ORIENTED FILE REGISTER OPERATIONS BCF f, b Bit Clear f 1 01 00bb bfff ffff 1, 2 BSF f, b Bit Set f 1 01 01bb bfff ffff 1, 2 BTFSC f, b Bit Test f, Skip if Clear 1 01 10bb bfff ffff 3 BTFSS f, b Bit Test f, Skip if Set 1 1 11 11bb bfff ffff 3 LITERAL AND CONTROL OPERATIONS ADDLW k Add literal and W 1 11 111 111x kkkk kkkk Z ANDLW k Add literal and W 1 11 11001 kkkk Z	XORWF	f, d	Exclusive OR W with f	1	00	0110	dfff	ffff	Z	1, 2
BCFf, bBit Clear f10100bbbfffffff1, 2BSFf, bBit Set f10101bbbfffffff1, 2BTFSCf, bBit Test f, Skip if Clear110110bbbfffffff1, 2BTFSSf, bBit Test f, Skip if Set110110bbbfffffff3LITERAL AND CONTROL OPERATIONSADDLWkAdd literal and W111111xkkk kkkkZANDLWkAdd literal with W1111001kkk kkkkZCALLkCall Subroutine2100kkk kkkkZCLRWDT-Clear Watchdog Timer10000000100TO, PDGOTOkGo to address2101kkk kkkkZIORLWkInclusive OR literal with W1111000 kkkk kkkkZMOVLWkMove literal to W111000x kkk kkkkZRETFIE-Return from interrupt20000001001RETURN-Return from Subroutine20000000001TO, PDSUBLWkSubtract w from literal111110x kkkk kkkkZVORLWkExclusive OR literal with W1111010 kkkk kkkkZ			BIT-ORIENTED FILE REGIS		RATION	IS				
BSFf, bBit Set f10101bbbfffffff1, 2BTFSCf, bBit Test f, Skip if Clear11110bbbfffffff3BTFSSf, bBit Test f, Skip if Set1(2)0110bbbfffffff3LITERAL AND CONTROL OPERATIONSADDLWkAdd literal and W111111111xkkkkkkkC, DC, ZANDLWkAND literal with W1111001kkkkkkkkZCALLkCall Subroutine2100kkkkkkkZCLRWDT-Clear Watchdog Timer10000000100TO, PDGOTOkGo to address2101kkkkkkkZIORLWkInclusive OR literal with W1111000kkkkKkkNOVLWkMove literal to W11101xxkkkkZRETFIE-Return from interrupt20000001001TO, PDRETURN-Return from Subroutine20000001001TO, PDSUBLWkSubtract w from literal111110xkkkkC, DC, ZXORLWkExclusive OR literal with W1111100kkkkZ	BCF	f. b	Bit Clear f	1	01	00bb	bfff	ffff		1.2
BTFSC BTFSSf, bBit Test f, Skip if Clear Bit Test f, Skip if Set1 (2) 1 (2)011 0bbbfffffff3LITERAL AND CONTROL OPERATIONSADDLW ANDLW CALLkAdd literal and W AND literal with W111111111xkkkkkkkkC, DC, Z 	BSF	f. b	Bit Set f	1	01	01bb	bfff	ffff		1.2
BTFSSf, bBit Test f, Skip if Set1 (2)0111bbbfffffff3LITERAL AND CONTROL OPERATIONSADDLWkAdd literal and W111111111xkkkkkkkkC, DC, ZANDLWkAND literal with W1111111xkkkkkkkkZCALLkCall Subroutine2100kkkkkkkkkkkZCLRWDT-Clear Watchdog Timer100000001100100TO, PDGOTOkGo to address2101kkkkkkkkkkkZIORLWkInclusive OR literal with W1111000kkkk kkkkZMOVLWkMove literal to W1110000000001001RETFIE-Return from interrupt200000000001001RETURN-Return from Subroutine20000000101TO, PDSUBLWkSubtract W from literal1111100kkkk kkkkZVORLWkSubtract W from literal11111010kkkk kkkkZ	BTFSC	f. b	Bit Test f. Skip if Clear	1 (2)	01	10bb	bfff	ffff		3
LITERAL AND CONTROL OPERATIONS ADDLW k Add literal and W 1 11 111 111x kkkk kkkk C, DC, Z ANDLW k AND literal with W 1 11 111 111x kkkk kkkk C, DC, Z CALL k Call Subroutine 2 10 0kkk kkkk kkkk Z CLRWDT - Clear Watchdog Timer 1 00 0000 0110 0100 TO, PD GOTO k Go to address 2 10 1kkk kkkk kkkk Z IORLW k Inclusive OR literal with W 1 11 1000 kkkk kkkk Z MOVLW k Inclusive OR literal with W 1 11 000xx kkkk Z MOVLW k Move literal to W 1 11 00xx kkkk Z MOVLW k Return from interrupt 2 00 0000 0001 001 RETURN Return from Subroutine 2 00 0000	BTFSS	f, b	Bit Test f, Skip if Set	1 (2)	01	11bb	bfff	ffff		3
ADDLWkAdd literal and W111111 111xkkkkkkkkC, DC, ZANDLWkAND literal with W1111001kkkkkkkkZCALLkCall Subroutine2100kkkkkkkkkkkZCLRWDT-Clear Watchdog Timer100000001100100TO, PDGOTOkGo to address2101kkkkkkkkkkkZIORLWkInclusive OR literal with W1111000kkkkkkkZMOVLWkMove literal to W11100xxkkkkkkkZRETFIE-Return from interrupt20000001001TO, PDRETURN-Return from Subroutine20000000001TO, PDSLEEP-Go into Standby mode111110xkkkkkkkkZSUBLWkSubtract W from literal111110xkkkkKkkkZ			LITERAL AND CONTRO		IONS				I	I
ANDLWkAND literal with W1111011klkkklkkZANDLWkAND literal with W1111001kkkkkkZCALLkCall Subroutine2100kkkkkkkkkkZCLRWDT-Clear Watchdog Timer100000001100100GOTOkGo to address2101kkkkkkkkkkZIORLWkInclusive OR literal with W1111000kkkkkkkkZMOVLWkMove literal to W11100xxkkkkkkkkZRETFIE-Return from interrupt20000000001IRETURN-Return from Subroutine20000000100ISLEEP-Go into Standby mode1111100xkkkkkkkkC, DC, ZSUBLWkExclusive OR literal with W1111010xkkkkKkkkZ		k	Add literal and W	1	11	111x	kkkk	kkkk	C DC Z	
CALLkCall Subroutine2100kkkkkkkTO, PDCLRWDT-Clear Watchdog Timer100000001100100TO, PDGOTOkGo to address2101kkkkkkkkkkkZIORLWkInclusive OR literal with W1111000kkkkkkkkZMOVLWkMove literal to W11100xxkkkkkkkkZRETFIE-Return from interrupt20000000001IRETURN-Return with literal in W21101xxkkkkkkkkRETURN-Return from Subroutine20000000101TO, PDSUBLWkSubtract W from literal111110xkkkkkkkkZXORLWkExclusive OR literal with W1111010kkkkZ	ANDLW	k	AND literal with W	1	11	1001	kkkk	kkkk	Z	
CLRWDT – Clear Watchdog Timer 1 00 0000 0110 0100 GOTO k Go to address 2 10 1kkk kkkk kkkk IORLW k Inclusive OR literal with W 1 11 1000 kkkk kkkk IORLW k Inclusive OR literal with W 1 11 1000 kkkk kkkk MOVLW k Move literal to W 1 11 00000 0000 1001 RETFIE – Return from interrupt 2 00 00000 1001 RETLW k Return from Subroutine 2 00 0000 1000 SLEEP – Go into Standby mode 1 00 0000 011 TO, PD SUBLW k Subtract W from literal 1 11 1100 kkkk Kkkk Z, DC, Z XORLW k Exclusive OR literal with W 1 11 1010 kkkk Z	CALL	k	Call Subroutine	2	10	0kkk	kkkk	kkkk	_	
GOTO k Go to address 2 10 1kkk kkk kkk IORLW k Inclusive OR literal with W 1 11 1000 kkk kkk kkk Z MOVLW k Move literal to W 1 11 1000 kkk kkk kkk Z MOVLW k Move literal to W 1 11 00xx kkk kkk k RETFIE - Return from interrupt 2 00 0000 0001 Image: state	CLRWDT	_	Clear Watchdog Timer	1	0.0	0000	0110	0100	TO PD	
IORLW k Inclusive OR literal with W 1 11 1000 kkk kkk Z MOVLW k Move literal to W 1 11 1000 kkk kkk Z MOVLW k Move literal to W 1 11 1000 kkk kkk Z RETFIE - Return from interrupt 2 00 0000 0001 R RETLW k Return with literal in W 2 11 01xx kkk kkk RETURN - Return from Subroutine 2 00 0000 0001 TO, PD SUBLW k Subtract W from literal 1 11 1100x kkk kkkk Z, DC, Z XORLW k Exclusive OR literal with W 1 11 1010 kkkk Z, DC, Z	GOTO	k	Go to address	2	10	1kkk	kkkk	kkkk		
MOVLW k Move literal to W 1 11 100xx kkk kkkk RETFIE - Return from interrupt 2 00 0000 0001 1001 RETLW k Return with literal in W 2 11 01xx kkkk kkkk RETURN - Return from Subroutine 2 00 0000 0001 100 SLEEP - Go into Standby mode 1 01 100 0011 TO, PD SUBLW k Subtract W from literal 1 11 110x kkkk kkkk Z, DC, Z XORLW k Exclusive OR literal with W 1 11 1010 kkkk Z, DC, Z	IORI W	k	Inclusive OR literal with W	1	11	1000	kkkk	kkkk	7	
RETFIE - Return from interrupt 2 00 0000 0000 1001 RETFIE - Return from interrupt 2 11 01xx kkkk kkkk RETURN - Return from Subroutine 2 00 0000 1000 SLEEP - Go into Standby mode 1 00 0000 011 TO, PD SUBLW k Subtract W from literal 1 11 110x kkkk kkkk Z, DC, Z	MOVIW	k	Move literal to W	1	11	00xx	kkkk	kkkk	-	
RETLW k Return with literal in W 2 11 01xx kkkk kkkk RETURN - Return from Subroutine 2 00 0000 1000 SLEEP - Go into Standby mode 1 00 0000 011 TO, PD SUBLW k Subtract w from literal 1 11 110x kkkk kkkk Z	RETFIE	_	Return from interrupt	2	00	0000	0000	1001		
RETURN Return from Subroutine 2 11 0100 1000 SLEEP - Go into Standby mode 1 00 0000 0100 1 SUBLW k Subtract W from literal 1 11 110x kkkk kkkk C, DC, Z XORLW k Exclusive OR literal with W 1 11 1010 kkkk kkkk Z	RETLW	k	Return with literal in W	2	11	01xx	kkkk	kkkk		
SLEEP – Go into Standby mode 1 00 0000 0110 0011 TO, PD SUBLW k Subtract W from literal 1 11 110x kkkk kkkk C, DC, Z XORLW k Exclusive OR literal with W 1 11 1010 kkkk Z	RETURN	_	Return from Subroutine	2	0.0	0000	0000	1000		
SUBLW k Subtract w from literal 1 11 110x kkkk K, DC, Z XORLW k Exclusive OR literal with W 1 11 1010 kkkk Z	SLEEP	_	Go into Standby mode	1	00	0000	0110	0011	TO, PD	
XORLW k Exclusive OR literal with W 1 11 1010 kkkk kkkk 7	SUBLW	k	Subtract w from literal	1	11	110×	kkkk	kkkk	C. DC. 7	
	XORLW	k	Exclusive OR literal with W	1	11	1010	kkkk	kkkk	Z, 2, 2, 2	

TABLE 15-2: PIC16F883/884/886/887 INSTRUCTION SET

Note 1: When an I/O register is modified as a function of itself (e.g., MOVF GPIO, 1), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and where applicable, d = 1), the prescaler will be cleared if assigned to the Timer0 module.

3: If the Program Counter (PC) is modified, or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

ADDLW	Add literal and W
Syntax:	[<i>label</i>] ADDLW k
Operands:	$0 \le k \le 255$
Operation:	$(W) + k \to (W)$
Status Affected:	C, DC, Z
Description:	The contents of the W register are added to the eight-bit literal 'k and the result is placed in the W register.

BCF	Bit Clear f
Syntax:	[label]BCF f,b
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b \leq 7 \end{array}$
Operation:	$0 \rightarrow (f < b >)$
Status Affected:	None
Description:	Bit 'b' in register 'f' is cleared.

ADDWF	Add W and f
Syntax:	[label] ADDWF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(W) + (f) \rightarrow (destination)
Status Affected:	C, DC, Z
Description:	Add the contents of the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

Bit Set f
[<i>label</i>] BSF f,b
$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b \leq 7 \end{array}$
$1 \rightarrow (f < b >)$
None
Bit 'b' in register 'f' is set.

ANDLW	AND literal with W
Syntax:	[label] ANDLW k
Operands:	$0 \le k \le 255$
Operation:	(W) .AND. (k) \rightarrow (W)
Status Affected:	Z
Description:	The contents of W register are AND'ed with the eight-bit literal 'k'. The result is placed in the W register.

BTFSC	Bit Test f, Skip if Clear
Syntax:	[label] BTFSC f,b
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b \leq 7 \end{array}$
Operation:	skip if (f) = 0
Status Affected:	None
Description:	If bit 'b' in register 'f' is '1', the next instruction is executed. If bit 'b' in register 'f' is '0', the next instruction is discarded, and a NOP is executed instead, making this a two-cycle instruction.

ANDWF	AND W with f
Syntax:	[<i>label</i>] ANDWF f,d
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	(W) .AND. (f) \rightarrow (destination)
Status Affected:	Z
Description:	AND the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

Instruction Descriptions

15.2

BTFSS	Bit Test f, Skip if Set
Syntax:	[label] BTFSS f,b
Operands:	$0 \le f \le 127$ $0 \le b < 7$
Operation:	skip if (f) = 1
Status Affected:	None
Description:	If bit 'b' in register 'f' is '0', the next instruction is executed. If bit 'b' is '1', then the next instruction is discarded and a NOP is executed instead, making this a two-cycle instruction.

CLRWDT	Clear Watchdog Timer
Syntax:	[label] CLRWDT
Operands:	None
Operation: Status Affected:	$\begin{array}{l} 00h \rightarrow WDT \\ 0 \rightarrow WDT \ prescaler, \\ 1 \rightarrow \overline{TO} \\ 1 \rightarrow \overline{PD} \\ \hline \overline{TO}, \ \overline{PD} \end{array}$
Description:	CLRWDT instruction resets the Watchdog Timer. It also resets the prescaler of the WDT. Status bits TO and PD are set.

CALL	Call Subroutine
Syntax:	[<i>label</i>] CALL k
Operands:	$0 \le k \le 2047$
Operation:	$\begin{array}{l} (PC)+1 \rightarrow TOS, \\ k \rightarrow PC < 10:0>, \\ (PCLATH < 4:3>) \rightarrow PC < 12:11> \end{array}$
Status Affected:	None
Description:	Call Subroutine. First, return address (PC + 1) is pushed onto the stack. The eleven-bit immediate address is loaded into PC bits <10:0>. The upper bits of the PC are loaded from PCLATH. CALL is a two-cycle instruction.

COMF	Complement f
Syntax:	[<i>label</i>] COMF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	$(\overline{f}) \rightarrow (destination)$
Status Affected:	Z
Description:	The contents of register 'f' are complemented. If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'.

CLRF	Clear f
Syntax:	[label] CLRF f
Operands:	$0 \le f \le 127$
Operation:	$\begin{array}{l} 00h \rightarrow (f) \\ 1 \rightarrow Z \end{array}$
Status Affected:	Z
Description:	The contents of register 'f' are cleared and the Z bit is set.

DECF	Decrement f
Syntax:	[label] DECF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(f) - 1 \rightarrow (destination)
Status Affected:	Z
Description:	Decrement register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

CLRW	Clear W
Syntax:	[label] CLRW
Operands:	None
Operation:	$\begin{array}{l} \text{O0h} \rightarrow (\text{W}) \\ \text{1} \rightarrow \text{Z} \end{array}$
Status Affected:	Z
Description:	W register is cleared. Zero bit (Z) is set.

DECFSZ	Decrement f, Skip if 0
Syntax:	[label] DECFSZ f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in \ [0,1] \end{array}$
Operation:	(f) - 1 \rightarrow (destination); skip if result = 0
Status Affected:	None
Description:	The contents of register 'f' are decremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'. If the result is '1', the next instruction is executed. If the result is '0', then a NOP is executed instead, making it a two-cycle instruction.

INCFSZ	Increment f, Skip if 0
Syntax:	[label] INCFSZ f,d
Operands:	0 ≤ f ≤ 127 d ∈ [0,1]
Operation:	(f) + 1 \rightarrow (destination), skip if result = 0
Status Affected:	None
Description:	The contents of register 'f' are incremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'. If the result is '1', the next instruction is executed. If the result is '0', a NOP is executed instead, making it a two-cycle instruction.

GOTO	Unconditional Branch
Syntax:	[<i>label</i>] GOTO k
Operands:	$0 \le k \le 2047$
Operation:	$\begin{array}{l} k \rightarrow PC <\!\! 10:\! 0 \!\!> \\ PCLATH <\!\! 4:\! 3 \!\!> \rightarrow PC <\!\! 12:\! 11 \!\!> \end{array}$
Status Affected:	None
Description:	GOTO is an unconditional branch. The eleven-bit immediate value is loaded into PC bits <10:0>. The upper bits of PC are loaded from PCLATH<4:3>. GOTO is a two-cycle instruction.

IORLW	Inclusive OR literal with W
Syntax:	[<i>label</i>] IORLW k
Operands:	$0 \le k \le 255$
Operation:	(W) .OR. $k \rightarrow$ (W)
Status Affected:	Z
Description:	The contents of the W register are OR'ed with the eight-bit literal 'k'. The result is placed in the W register.

INCF	Increment f
Syntax:	[label] INCF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(f) + 1 \rightarrow (destination)
Status Affected:	Z
Description:	The contents of register 'f' are incremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.

IORWF	Inclusive OR W with f
Syntax:	[<i>label</i>] IORWF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in \ensuremath{\left[0,1 \right]} \end{array}$
Operation:	(W) .OR. (f) \rightarrow (destination)
Status Affected:	Z
Description:	Inclusive OR the W register with register 'f'. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.

MOVF	Move f
Syntax:	[<i>label</i>] MOVF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in \ensuremath{\left[0,1\right]} \end{array}$
Operation:	$(f) \rightarrow (dest)$
Status Affected:	Z
Description:	The contents of register 'f' is moved to a destination dependent upon the status of 'd'. If $d = 0$, destination is W register. If $d = 1$, the destination is file register 'f' itself. $d = 1$ is useful to test a file register since status flag Z is affected.
Words:	1
Cycles:	1
Example:	MOVF FSR, 0
	After Instruction W = value in FSR register Z = 1

MOVWF	Move W to f
Syntax:	[label] MOVWF f
Operands:	$0 \le f \le 127$
Operation:	$(W) \rightarrow (f)$
Status Affected:	None
Description:	Move data from W register to register 'f'.
Words:	1
Cycles:	1
Example:	MOVW OPTION F
	Before Instruction
	OPTION = 0xFF
	VV = 0x4F
	OPTION = 0x4F
	W = 0x4F

MOVLW	Move literal to W
Syntax:	[<i>label</i>] MOVLW k
Operands:	$0 \le k \le 255$
Operation:	$k \rightarrow (W)$
Status Affected:	None
Description:	The eight-bit literal 'k' is loaded into W register. The "don't cares" will assemble as '0's.
Words:	1
Cycles:	1
Example:	MOVLW 0x5A
	After Instruction W = 0x5A

NOP	No Operation
Syntax:	[label] NOP
Operands:	None
Operation:	No operation
Status Affected:	None
Description:	No operation.
Words:	1
Cycles:	1
<u>Example</u> :	NOP

RETFIE	Return from Interrupt	RETLW	Return with literal in W
Syntax:	[label] RETFIE	Syntax:	[<i>label</i>] RETLW k
Operands:	None	Operands:	$0 \le k \le 255$
Operation:	$\begin{array}{l} TOS \to PC, \\ 1 \to GIE \end{array}$	Operation:	$\begin{array}{l} k \rightarrow (W);\\ TOS \rightarrow PC \end{array}$
Status Affected:	None	Status Affected:	None
Description: Return from Interrupt. Stack is POPed and Top-of-Stack (TOS) loaded in the PC. Interrupts are enabled by setting Global Interrupt Enable bit, GIE		Description:	The W register is loaded with the eight-bit literal 'k'. The program counter is loaded from the top of the stack (the return address). This is a two-cycle instruction.
	(INTCON<7>). This is a two-cycle	Words:	1
Wordo.		Cycles:	2
Cycles:	2	Example:	CALL TABLE;W contains table
Example:	RETFIE		;offset value
	After Interrupt PC = TOS GIE = 1	TABLE	<pre>• ;W now has • ;table value • • ADDWF PC ;W = offset</pre>

RETURN	Return from Subroutine		
Syntax:	[label] RETURN		
Operands:	None		
Operation:	$TOS\toPC$		
Status Affected:	None		
Description:	Return from subroutine. The stack is POPed and the top of the stack (TOS) is loaded into the program counter. This is a two-cycle instruction.		

RETLW k1 ;Begin table

RETLW kn ; End of table

W = 0x07

value of k8

RETLW k2 ;

Before Instruction

After Instruction W =

•

RLF	Rotate	Left f th	roug	h Carry	y
Syntax:	[label]	RLF	f,d		
Operands:	$0 \le f \le f$ $d \in [0, 2]$	127 1]			
Operation:	See de	scription	belov	N	
Status Affected:	С				
Description:	The con rotated the Car result is If 'd' is back in	ntents of one bit to rry flag. If s placed '1', the re register	regis o the f 'd' is in the esult i 'f'. Regis	ter 'f' a left thro s '0', the W reg is store ster f	re bugh sister. d
Words:	1				
Cycles:	1				
Example:	RLF	REG1	,0		
	Before Instruction				
		REG1	=	1110	0110
	A ft a r l -	C	=	0	
	After In	SILUCTION			0110
		REGI	=	1100	0110
		w	=	1	TTOO
		C	-	1	

SLEEP	Enter Sleep mode
Syntax:	[label] SLEEP
Operands:	None
Operation:	$\begin{array}{l} 00h \rightarrow WDT, \\ 0 \rightarrow \underline{WDT} \text{ prescaler}, \\ 1 \rightarrow \overline{TO}, \\ 0 \rightarrow \overline{PD} \end{array}$
Status Affected:	TO, PD
Description:	The power-down Status bit, \overline{PD} is cleared. Time-out Status bit, \overline{TO} is set. Watchdog Timer and its prescaler are cleared. The processor is put into Sleep mode with the oscillator stopped.

RRF	Rotate Right f through Carry	
Syntax:	[<i>label</i>] RRF f,d	
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in \ [0,1] \end{array}$	
Operation:	See description below	
Status Affected:	С	
Description:	The contents of register 'f' are rotated one bit to the right through the Carry flag. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.	
	C Register f	

SUBLW	Subtract W from literal			
Syntax:	[label] Sl	JBLW k		
Operands:	$0 \le k \le 255$			
Operation:	$k - (W) \to (W)$			
Status Affected:	C, DC, Z			
Description:	The W register is subtracted (2's complement method) from the eight-bit literal 'k'. The result is placed in the W register.			
	C = 0	W > k		
	C = 1	$W \leq k$		
	DC = 0	W<3:0> > k<3:0>		

DC = 1

 $W < 3:0 > \le k < 3:0 >$

SUBWF	Subtract W	from f	
Syntax:	[label] Sl	JBWF f,d	
Operands:	$0 \le f \le 127$ $d \in [0,1]$		
Operation:	(f) - (W) \rightarrow ((destination)	
Status Affected:	C, DC, Z		
Description:	Subtract (2's complement method) W register from register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.		
	C = 0	W > f	
	1		

$\mathbf{C} = 0$	VV > f
C = 1	$W \leq f$
DC = 0	W<3:0>>f<3:0>
DC = 1	$W < 3:0 > \le f < 3:0 >$

XORWF	Exclusive OR W with f	
Syntax:	[label] XORWF f,d	
Operands:	$0 \le f \le 127$ $d \in [0,1]$	
Operation:	(W) .XOR. (f) \rightarrow (destination)	
Status Affected:	Z	
Description:	Exclusive OR the contents of the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.	

SWAPF	Swap Nibbles in f
Syntax:	[label] SWAPF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	$(f < 3:0 >) \rightarrow (destination < 7:4 >),$ $(f < 7:4 >) \rightarrow (destination < 3:0 >)$
Status Affected:	None
Description:	The upper and lower nibbles of register 'f' are exchanged. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed in register 'f'.

XORLW	Exclusive OR literal with W
Syntax:	[<i>label</i>] XORLW k
Operands:	$0 \le k \le 255$
Operation:	(W) .XOR. $k \rightarrow (W)$
Status Affected:	Z
Description:	The contents of the W register are XOR'ed with the eight-bit literal 'k'. The result is placed in the W register.

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NOTES:

16.0 DEVELOPMENT SUPPORT

The PIC[®] microcontrollers are supported with a full range of hardware and software development tools:

- Integrated Development Environment
 - MPLAB[®] IDE Software
- Assemblers/Compilers/Linkers
 - MPASM[™] Assembler
 - MPLAB C18 and MPLAB C30 C Compilers
 - MPLINK[™] Object Linker/
 - MPLIB[™] Object Librarian
 - MPLAB ASM30 Assembler/Linker/Library
- Simulators
 - MPLAB SIM Software Simulator
- Emulators
 - MPLAB ICE 2000 In-Circuit Emulator
 - MPLAB REAL ICE™ In-Circuit Emulator
- In-Circuit Debugger
 - MPLAB ICD 2
- Device Programmers
 - PICSTART[®] Plus Development Programmer
 - MPLAB PM3 Device Programmer
 - PICkit[™] 2 Development Programmer
- Low-Cost Demonstration and Development Boards and Evaluation Kits

16.1 MPLAB Integrated Development Environment Software

The MPLAB IDE software brings an ease of software development previously unseen in the 8/16-bit microcontroller market. The MPLAB IDE is a Windows[®] operating system-based application that contains:

- A single graphical interface to all debugging tools
 - Simulator
 - Programmer (sold separately)
 - Emulator (sold separately)
 - In-Circuit Debugger (sold separately)
- · A full-featured editor with color-coded context
- A multiple project manager
- Customizable data windows with direct edit of contents
- High-level source code debugging
- Visual device initializer for easy register initialization
- Mouse over variable inspection
- Drag and drop variables from source to watch windows
- Extensive on-line help
- Integration of select third party tools, such as HI-TECH Software C Compilers and IAR C Compilers

The MPLAB IDE allows you to:

- Edit your source files (either assembly or C)
- One touch assemble (or compile) and download to PIC MCU emulator and simulator tools (automatically updates all project information)
- Debug using:
 - Source files (assembly or C)
 - Mixed assembly and C
 - Machine code

MPLAB IDE supports multiple debugging tools in a single development paradigm, from the cost-effective simulators, through low-cost in-circuit debuggers, to full-featured emulators. This eliminates the learning curve when upgrading to tools with increased flexibility and power.

16.2 MPASM Assembler

The MPASM Assembler is a full-featured, universal macro assembler for all PIC MCUs.

The MPASM Assembler generates relocatable object files for the MPLINK Object Linker, Intel[®] standard HEX files, MAP files to detail memory usage and symbol reference, absolute LST files that contain source lines and generated machine code and COFF files for debugging.

The MPASM Assembler features include:

- Integration into MPLAB IDE projects
- User-defined macros to streamline
 assembly code
- Conditional assembly for multi-purpose source files
- Directives that allow complete control over the assembly process

16.3 MPLAB C18 and MPLAB C30 C Compilers

The MPLAB C18 and MPLAB C30 Code Development Systems are complete ANSI C compilers for Microchip's PIC18 and PIC24 families of microcontrollers and the dsPIC30 and dsPIC33 family of digital signal controllers. These compilers provide powerful integration capabilities, superior code optimization and ease of use not found with other compilers.

For easy source level debugging, the compilers provide symbol information that is optimized to the MPLAB IDE debugger.

16.4 MPLINK Object Linker/ MPLIB Object Librarian

The MPLINK Object Linker combines relocatable objects created by the MPASM Assembler and the MPLAB C18 C Compiler. It can link relocatable objects from precompiled libraries, using directives from a linker script.

The MPLIB Object Librarian manages the creation and modification of library files of precompiled code. When a routine from a library is called from a source file, only the modules that contain that routine will be linked in with the application. This allows large libraries to be used efficiently in many different applications.

The object linker/library features include:

- Efficient linking of single libraries instead of many smaller files
- Enhanced code maintainability by grouping related modules together
- Flexible creation of libraries with easy module listing, replacement, deletion and extraction

16.5 MPLAB ASM30 Assembler, Linker and Librarian

MPLAB ASM30 Assembler produces relocatable machine code from symbolic assembly language for dsPIC30F devices. MPLAB C30 C Compiler uses the assembler to produce its object file. The assembler generates relocatable object files that can then be archived or linked with other relocatable object files and archives to create an executable file. Notable features of the assembler include:

- Support for the entire dsPIC30F instruction set
- Support for fixed-point and floating-point data
- Command line interface
- Rich directive set
- Flexible macro language
- MPLAB IDE compatibility

16.6 MPLAB SIM Software Simulator

The MPLAB SIM Software Simulator allows code development in a PC-hosted environment by simulating the PIC MCUs and dsPIC[®] DSCs on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a comprehensive stimulus controller. Registers can be logged to files for further run-time analysis. The trace buffer and logic analyzer display extend the power of the simulator to record and track program execution, actions on I/O, most peripherals and internal registers.

The MPLAB SIM Software Simulator fully supports symbolic debugging using the MPLAB C18 and MPLAB C30 C Compilers, and the MPASM and MPLAB ASM30 Assemblers. The software simulator offers the flexibility to develop and debug code outside of the hardware laboratory environment, making it an excellent, economical software development tool.

16.7 MPLAB ICE 2000 High-Performance In-Circuit Emulator

The MPLAB ICE 2000 In-Circuit Emulator is intended to provide the product development engineer with a complete microcontroller design tool set for PIC microcontrollers. Software control of the MPLAB ICE 2000 In-Circuit Emulator is advanced by the MPLAB Integrated Development Environment, which allows editing, building, downloading and source debugging from a single environment.

The MPLAB ICE 2000 is a full-featured emulator system with enhanced trace, trigger and data monitoring features. Interchangeable processor modules allow the system to be easily reconfigured for emulation of different processors. The architecture of the MPLAB ICE 2000 In-Circuit Emulator allows expansion to support new PIC microcontrollers.

The MPLAB ICE 2000 In-Circuit Emulator system has been designed as a real-time emulation system with advanced features that are typically found on more expensive development tools. The PC platform and Microsoft[®] Windows[®] 32-bit operating system were chosen to best make these features available in a simple, unified application.

16.8 MPLAB REAL ICE In-Circuit Emulator System

MPLAB REAL ICE In-Circuit Emulator System is Microchip's next generation high-speed emulator for Microchip Flash DSC[®] and MCU devices. It debugs and programs PIC[®] and dsPIC[®] Flash microcontrollers with the easy-to-use, powerful graphical user interface of the MPLAB Integrated Development Environment (IDE), included with each kit.

The MPLAB REAL ICE probe is connected to the design engineer's PC using a high-speed USB 2.0 interface and is connected to the target with either a connector compatible with the popular MPLAB ICD 2 system (RJ11) or with the new high speed, noise tolerant, lowvoltage differential signal (LVDS) interconnection (CAT5).

MPLAB REAL ICE is field upgradeable through future firmware downloads in MPLAB IDE. In upcoming releases of MPLAB IDE, new devices will be supported, and new features will be added, such as software breakpoints and assembly code trace. MPLAB REAL ICE offers significant advantages over competitive emulators including low-cost, full-speed emulation, real-time variable watches, trace analysis, complex breakpoints, a ruggedized probe interface and long (up to three meters) interconnection cables.

16.9 MPLAB ICD 2 In-Circuit Debugger

Microchip's In-Circuit Debugger, MPLAB ICD 2, is a powerful, low-cost, run-time development tool, connecting to the host PC via an RS-232 or high-speed USB interface. This tool is based on the Flash PIC MCUs and can be used to develop for these and other PIC MCUs and dsPIC DSCs. The MPLAB ICD 2 utilizes the in-circuit debugging capability built into the Flash devices. This feature, along with Microchip's In-Circuit Serial Programming[™] (ICSP[™]) protocol, offers costeffective, in-circuit Flash debugging from the graphical user interface of the MPLAB Integrated Development Environment. This enables a designer to develop and debug source code by setting breakpoints, single stepping and watching variables, and CPU status and peripheral registers. Running at full speed enables testing hardware and applications in real time. MPLAB ICD 2 also serves as a development programmer for selected PIC devices.

16.10 MPLAB PM3 Device Programmer

The MPLAB PM3 Device Programmer is a universal, CE compliant device programmer with programmable voltage verification at VDDMIN and VDDMAX for maximum reliability. It features a large LCD display (128 x 64) for menus and error messages and a modular, detachable socket assembly to support various package types. The ICSP™ cable assembly is included as a standard item. In Stand-Alone mode, the MPLAB PM3 Device Programmer can read, verify and program PIC devices without a PC connection. It can also set code protection in this mode. The MPLAB PM3 connects to the host PC via an RS-232 or USB cable. The MPLAB PM3 has high-speed communications and optimized algorithms for quick programming of large memory devices and incorporates an SD/MMC card for file storage and secure data applications.

16.11 PICSTART Plus Development Programmer

The PICSTART Plus Development Programmer is an easy-to-use, low-cost, prototype programmer. It connects to the PC via a COM (RS-232) port. MPLAB Integrated Development Environment software makes using the programmer simple and efficient. The PICSTART Plus Development Programmer supports most PIC devices in DIP packages up to 40 pins. Larger pin count devices, such as the PIC16C92X and PIC17C76X, may be supported with an adapter socket. The PICSTART Plus Development Programmer is CE compliant.

16.12 PICkit 2 Development Programmer

The PICkit[™] 2 Development Programmer is a low-cost programmer and selected Flash device debugger with an easy-to-use interface for programming many of Microchip's baseline, mid-range and PIC18F families of Flash memory microcontrollers. The PICkit 2 Starter Kit includes a prototyping development board, twelve sequential lessons, software and HI-TECH's PICC[™] Lite C compiler, and is designed to help get up to speed quickly using PIC[®] microcontrollers. The kit provides everything needed to program, evaluate and develop applications using Microchip's powerful, mid-range Flash memory family of microcontrollers.

16.13 Demonstration, Development and Evaluation Boards

A wide variety of demonstration, development and evaluation boards for various PIC MCUs and dsPIC DSCs allows quick application development on fully functional systems. Most boards include prototyping areas for adding custom circuitry and provide application firmware and source code for examination and modification.

The boards support a variety of features, including LEDs, temperature sensors, switches, speakers, RS-232 interfaces, LCD displays, potentiometers and additional EEPROM memory.

The demonstration and development boards can be used in teaching environments, for prototyping custom circuits and for learning about various microcontroller applications.

In addition to the PICDEM[™] and dsPICDEM[™] demonstration/development board series of circuits, Microchip has a line of evaluation kits and demonstration software for analog filter design, KEELOQ[®] security ICs, CAN, IrDA[®], PowerSmart[®] battery management, SEEVAL[®] evaluation system, Sigma-Delta ADC, flow rate sensing, plus many more.

Check the Microchip web page (www.microchip.com) and the latest *"Product Selector Guide"* (DS00148) for the complete list of demonstration, development and evaluation kits.

17.0 ELECTRICAL SPECIFICATIONS

Absolute Maximum Ratings^(†)

Ambient temperature under bias	40° to +125°C
Storage temperature	65°C to +150°C
Voltage on VDD with respect to VSS	0.3V to +6.5V
Voltage on MCLR with respect to Vss	0.3V to +13.5V
Voltage on all other pins with respect to Vss	0.3V to (VDD + 0.3V)
Total power dissipation ⁽¹⁾	800 mW
Maximum current out of Vss pin	300 mA
Maximum current into VDD pin	250 mA
Input clamp current, Iik (VI < 0 or VI > VDD)	± 20 mA
Output clamp current, Ioк (Vo < 0 or Vo >VDD)	± 20 mA
Maximum output current sunk by any I/O pin	25 mA
Maximum output current sourced by any I/O pin	
Maximum current sunk by PORTA, PORTB and PORTE (combined) ⁽²⁾	200 mA
Maximum current sourced by PORTA, PORTB and PORTE (combined) ⁽²⁾	200 mA
Maximum current sunk by PORTC and PORTD (combined) ⁽²⁾	200 mA
Maximum current sourced by PORTC and PORTD (combined) ⁽²⁾	200 mA
Note 1: Power dissipation is calculated as follows: PDIS = VDD x {IDD $-\sum$ IOH} + \sum {(VDD -2 : PORTD and PORTE are implemented on PIC16F886/PIC16F887 only.	- VOH) x IOH} + Σ (VOI x IOL).

† NOTICE: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure above maximum rating conditions for extended periods may affect device reliability.





FIGURE 17-2: HFINTOSC FREQUENCY ACCURACY OVER DEVICE VDD AND TEMPERATURE



17.1 DC Characteristics: PIC16F882/883/884/886/887-I (Industrial) PIC16F882/883/884/886/887-E (Extended)

DC CHARACTERISTICS				$\begin{array}{ll} \mbox{Standard Operating Conditions (unless otherwise stated)} \\ \mbox{Operating temperature} & -40^{\circ}C \leq TA \leq +85^{\circ}C \mbox{ for industrial} \\ & -40^{\circ}C \leq TA \leq +125^{\circ}C \mbox{ for extended} \end{array}$									
Param No.	Sym	Characteristic	Min Typ† Max Units Conditions										
	Vdd	Supply Voltage	2.0	—	5.5	V	Fosc < = 8 MHz: HFINTOSC, EC						
D001			2.0		5.5	V	Fosc < = 4 MHz						
D001C			3.0		5.5	V	Fosc < = 10 MHz						
D001D			4.5	—	5.5	V	Fosc < = 20 MHz						
D002*	Vdr	RAM Data Retention Voltage ⁽¹⁾	1.5	_		V	Device in Sleep mode						
D003	VPOR	VDD Start Voltage to ensure internal Power-on Reset signal	—	Vss	_	V	See Section 14.2.1 "Power-on Reset (POR)" for details.						
D004* SVDD VDD Rise Rate to ensure internal Power-on Reset signal 0.05 — — V/ms See Section 14.2.1 "Power-on (POR)" for details.													
ŕ	* These	parameters are characterize	d but n	ot test	ed								

These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered in Sleep mode without losing RAM data.

17.2 DC Characteristics: PIC16F882/883/884/886/887-I (Industrial) PIC16F882/883/884/886/887-E (Extended)

DC CHA	ARACTERISTICS	Standa Operat	ard Operating temp	ating Co erature	onditions ≥ 40°C- -40°C-	$\begin{array}{l} \textbf{-ditions (unless otherwise stated)} \\ \textbf{-40^{\circ}C} \leq TA \leq \textbf{+85^{\circ}C} \text{ for industrial} \\ \textbf{-40^{\circ}C} \leq TA \leq \textbf{+125^{\circ}C} \text{ for extended} \end{array}$				
Param	Device Characteristics	Min	Trunt	Max	Unite	Conditions				
No.	Device Characteristics	IVIIN	турт	wax	Units	Vdd	Note			
D010	Supply Current (IDD) ^(1, 2)		11	23	μA	2.0	Fosc = 32 kHz			
			18	38	μA	3.0	LP Oscillator mode			
			35	75	μA	5.0				
D011*			140	250	μA	2.0	Fosc = 1 MHz			
			220	400	μA	3.0	XT Oscillator mode			
			380	650	μA	5.0				
D012			260	380	μA	2.0	Fosc = 4 MHz			
			420	670	μA	3.0	XT Oscillator mode			
			0.8	1.4	mA	5.0				
D013*		_	130	220	μA	2.0	Fosc = 1 MHz			
			215	360	μA	3.0	EC Oscillator mode			
			360	520	μA	5.0				
D014		_	220	340	μA	2.0	Fosc = 4 MHz			
		_	375	550	μA	3.0	EC Oscillator mode			
			0.65	1.0	mA	5.0				
D015		_	8	20	μA	2.0	Fosc = 31 kHz			
		_	16	40	μA	3.0	LFINTOSC mode			
			31	65	μA	5.0				
D016*		_	340	400	μA	2.0	Fosc = 4 MHz			
			500	650	μA	3.0	HFINTOSC mode			
			0.8	1.2	mA	5.0				
D017			410	0.7	μA	2.0	Fosc = 8 MHz			
			700	1	μA	3.0	HFINTOSC mode			
			1.30	1.8	mA	5.0				
D018		—	230	580	μA	2.0	Fosc = 4 MHz			
		_	400	950	μA	3.0	EXTRC mode ⁽³⁾			
		_	0.63	1.6	mA	5.0				
D019		—	2.6	3.7	mA	4.5	Fosc = 20 MHz			
			2.8	3.8	mA	5.0	HS Oscillator mode			

These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The test conditions for all IDD measurements in active operation mode are: OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD; MCLR = VDD; WDT disabled.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors, such as I/O pin loading and switching rate, oscillator type, internal code execution pattern and temperature, also have an impact on the current consumption.

3: For RC oscillator configurations, current through REXT is not included. The current through the resistor can be extended by the formula IR = VDD/2REXT (mA) with REXT in k Ω .

DC CHARACTERISTICS		Standa Operat	ard Oper ting temp	ating Co erature	onditions -40°C	s (unless otherwise stated) $S \leq TA \leq +85^{\circ}C$ for industrial			
Param	Device Characteristics	Min	Turnet	Max	Unito		Conditions		
No.	Device Characteristics	WIN	турт	wax	Units	Vdd	Note		
D020	Power-down Base	_	0.05	1.2	μA	2.0	WDT, BOR, Comparators, VREF and		
	Current(IPD) ⁽²⁾	—	0.15	1.5	μA	3.0	T1OSC disabled		
		—	0.35	1.8	μA	5.0			
		—	150	500	nA	3.0	$-40^{\circ}C \le TA \le +25^{\circ}C$		
D021		—	1.0	2.2	μA	2.0	WDT Current ⁽¹⁾		
		—	2.0	4.0	μA	3.0			
		—	3.0	7.0	μΑ	5.0			
D022		—	42	60	μA	3.0	BOR Current ⁽¹⁾		
		—	85	122	μΑ	5.0			
D023		—	32	45	μΑ	2.0	Comparator Current ⁽¹⁾ , both		
		—	60	78	μΑ	3.0	comparators enabled		
		—	120	160	μΑ	5.0			
D024		—	30	36	μΑ	2.0	CVREF Current ⁽¹⁾ (high range)		
		—	45	55	μΑ	3.0			
		—	75	95	μΑ	5.0			
D025*		_	39	47	μΑ	2.0	CVREF Current ⁽¹⁾ (low range)		
		—	59	72	μΑ	3.0			
		—	98	124	μΑ	5.0			
D026		_	4.5	7.0	μA	2.0	T1OSC Current ⁽¹⁾ , 32.768 kHz		
			5.0	8.0	μΑ	3.0			
		_	6.0	12	μΑ	5.0			
D027		_	0.30	1.6	μΑ	3.0	A/D Current ⁽¹⁾ , no conversion in		
		—	0.36	1.9	μA	5.0	progress		

17.3 DC Characteristics: PIC16F882/883/884/886/887-I (Industrial)

These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The peripheral current is the sum of the base IDD or IPD and the additional current consumed when this peripheral is enabled. The peripheral Δ current can be determined by subtracting the base IDD or IPD current from this limit. Max values should be used when calculating total current consumption.

2: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD.

17.4 DC Characteristics: PIC16F882/883/884/886/887-E (Extended)

DC CHA	DC CHARACTERISTICS		ard Oper ing temp	ating Co erature	onditions -40°C :	nditions (unless otherwise stated) $-40^{\circ}C \le TA \le +125^{\circ}C$ for extended			
Param	Device Characteristics	Min	Trunt	Max	Unite		Conditions		
No.	Device Characteristics	WIIN	турт	wax	Units	Vdd	Note		
D020E	Power-down Base		0.05	9	μA	2.0	WDT, BOR, Comparators, VREF and		
	Current (IPD) ⁽²⁾		0.15	11	μA	3.0	T1OSC disabled		
			0.35	15	μA	5.0			
D021E			1	28	μΑ	2.0	WDT Current ⁽¹⁾		
			2	30	μΑ	3.0			
			3	35	μΑ	5.0			
D022E		_	42	65	μΑ	3.0	BOR Current ⁽¹⁾		
		_	85	127	μΑ	5.0			
D023E			32	45	μA	2.0	Comparator Current ⁽¹⁾ , both		
		_	60	78	μA	3.0	comparators enabled		
			120	160	μA	5.0			
D024E			30	70	μA	2.0	CVREF Current ⁽¹⁾ (high range)		
		_	45	90	μΑ	3.0			
		—	75	120	μΑ	5.0			
D025E*		_	39	91	μΑ	2.0	CVREF Current ⁽¹⁾ (low range)		
			59	117	μΑ	3.0			
		—	98	156	μΑ	5.0			
D026E			4.5	25	μΑ	2.0	T1OSC Current ⁽¹⁾ , 32.768 kHz		
			5	30	μΑ	3.0			
		—	6	40	μA	5.0			
D027E			0.30	12	μA	3.0	A/D Current ⁽¹⁾ , no conversion in		
		—	0.36	16	μA	5.0	progress		

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The peripheral current is the sum of the base IDD or IPD and the additional current consumed when this peripheral is enabled. The peripheral Δ current can be determined by subtracting the base IDD or IPD current from this limit. Max values should be used when calculating total current consumption.

2: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD.

17.5 DC Characteristics: PIC16F882/883/884/886/887-I (Industrial) PIC16F882/883/884/886/887-E (Extended)

DC CHARACTERISTICS			Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial $-40^{\circ}C \le TA \le +125^{\circ}C$ for extended						
Param No.	Sym	Characteristic	Min Typ†		Мах	Units	Conditions		
	VIL	Input Low Voltage							
		I/O Port:							
D030		with TTL buffer	Vss	—	0.8	V	$4.5V \le VDD \le 5.5V$		
D030A			Vss	—	0.15 Vdd	V	$2.0V \le VDD \le 4.5V$		
D031		with Schmitt Trigger buffer	Vss	—	0.2 Vdd	V	$2.0V \le VDD \le 5.5V$		
D032		MCLR, OSC1 (RC mode) ⁽¹⁾	Vss	—	0.2 Vdd	V			
D033		OSC1 (XT and LP modes)	Vss	—	0.3	V			
D033A		OSC1 (HS mode)	Vss	_	0.3 Vdd	V			
	Viн	Input High Voltage							
		I/O ports:		—					
D040		with TTL buffer	2.0	—	Vdd	V	$4.5V \le VDD \le 5.5V$		
D040A			0.25 VDD + 0.8	—	Vdd	V	$2.0V \le VDD \le 4.5V$		
D041		with Schmitt Trigger buffer	0.8 Vdd	—	Vdd	V	$2.0V \le VDD \le 5.5V$		
D042		MCLR	0.8 Vdd	—	Vdd	V			
D043		OSC1 (XT and LP modes)	1.6	—	Vdd	V			
D043A		OSC1 (HS mode)	0.7 Vdd	—	Vdd	V			
D043B		OSC1 (RC mode)	0.9 Vdd	_	Vdd	V	(Note 1)		
	lı∟	Input Leakage Current ⁽²⁾							
D060		I/O ports	_	±0.1	± 1	μA	Vss ≤ VPIN ≤ VDD, Pin at high-impedance		
D061		MCLR ⁽³⁾	_	± 0.1	± 5	μΑ	$VSS \leq VPIN \leq VDD$		
D063		OSC1	_	±0.1	± 5	μA	Vss \leq VPIN \leq VDD, XT, HS and LP oscillator configuration		
D070*	IPUR	PORTA Weak Pull-up Current	50	250	400	μΑ	VDD = 5.0V, VPIN = VSS		
	Vol	Output Low Voltage ⁽⁵⁾							
D080		I/O ports	—	_	0.6	V	IOL = 8.5 mA, VDD = 4.5V (Ind.)		
	Vон	Output High Voltage ⁽⁵⁾							
D090		I/O ports	VDD - 0.7	—	_	V	IOH = -3.0 mA, VDD = 4.5V (Ind.)		

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: In RC oscillator configuration, the OSC1/CLKIN pin is a Schmitt Trigger input. It is not recommended to use an external clock in RC mode.

2: Negative current is defined as current sourced by the pin.

3: The leakage current on the MCLR pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

4: See Section 10.3.1 "Using the Data EEPROM" for additional information.

5: Including OSC2 in CLKOUT mode.

17.5 DC Characteristics: PIC16F882/883/884/886/887-I (Industrial) PIC16F882/883/884/886/887-E (Extended) (Continued)

DC CHARACTERISTICS			Standard Oper Operating temp	ss otherwise stated) +85°C for industrial +125°C for extended			
Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions
D100	IULP	Ultra Low-Power Wake-Up Current	_	200	—	nA	See Application Note AN879, "Using the Microchip Ultra Low-Power Wake-up Module" (DS00879)
		Capacitive Loading Specs on Output Pins					
D101*	COSC2	OSC2 pin	_	-	15	pF	In XT, HS and LP modes when external clock is used to drive OSC1
D101A*	Сю	All I/O pins	—	_	50	pF	
		Data EEPROM Memory					
D120	ED	Byte Endurance	100K	1M	—	E/W	-40°C ≤ TA ≤ +85°C
D120A	ED	Byte Endurance	10K	100K		E/W	+85°C ≤ TA ≤ +125°C
D121	Vdrw	VDD for Read/Write	VMIN	_	5.5	V	Using EECON1 to read/write VMIN = Minimum operating voltage
D122	TDEW	Erase/Write Cycle Time	—	5	6	ms	
D123	TRETD	Characteristic Retention	40	-	—	Year	Provided no other specifications are violated
D124	Tref	Number of Total Erase/Write Cycles before Refresh ⁽⁴⁾	1M	10M	—	E/W	$-40^{\circ}C \le TA \le +85^{\circ}C$
		Program Flash Memory					
D130	Eр	Cell Endurance	10K	100K	—	E/W	$-40^{\circ}C \le TA \le +85^{\circ}C$
D130A	ED	Cell Endurance	1K	10K	—	E/W	$+85^{\circ}C \le TA \le +125^{\circ}C$
D131	Vpr	VDD for Read	VMIN	-	5.5	V	Vміn = Minimum operating voltage
D132	VPEW	VDD for Erase/Write	4.5	-	5.5	V	
D133	TPEW	Erase/Write cycle time	—	2	2.5	ms	
D134	TRETD	Characteristic Retention	40	_	_	Year	Provided no other specifications are violated

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: In RC oscillator configuration, the OSC1/CLKIN pin is a Schmitt Trigger input. It is not recommended to use an external clock in RC mode.

2: Negative current is defined as current sourced by the pin.

3: The leakage current on the MCLR pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

4: See Section 10.3.1 "Using the Data EEPROM" for additional information.

5: Including OSC2 in CLKOUT mode.

17.6 Thermal Considerations

Standard Operating	Standard Operating Conditions (unless otherwise stated) Operating temperature $-40^{\circ}C \le TA \le +125^{\circ}C$										
Param No.	Sym	Characteristic	Тур	Units	Conditions						
TH01	θја	Thermal Resistance	47.2	C/W	40-pin PDIP package						
		Junction to Ambient	24.4	C/W	44-pin QFN package						
			45.8	C/W	44-pin TQFP package						
			60.2	C/W	28-pin PDIP package						
			80.2	C/W	28-pin SOIC package						
			89.4	C/W	28-pin SSOP package						
			29	C/W	28-pin QFN package						
TH02	θJC	Thermal Resistance	24.7	C/W	40-pin PDIP package						
		Junction to Case	TBD	C/W	44-pin QFN package						
			14.5	C/W	44-pin TQFP package						
			29	C/W	28-pin PDIP package						
			23.8	C/W	28-pin SOIC package						
			23.9	C/W	28-pin SSOP package						
			TBD	C/W	28-pin QFN package						
TH03	TJ	Junction Temperature	150	С	For derated power calculations						
TH04	PD	Power Dissipation	—	W	PD = PINTERNAL + PI/O						
TH05	PINTERNAL	Internal Power Dissipation		W	PINTERNAL = IDD x VDD						
					(NOTE 1)						
TH06	Pi/o	I/O Power Dissipation	—	W	$PI/O = \Sigma (IOL * VOL) + \Sigma (IOH * (VDD - VOH))$						
TH07	Pder	Derated Power	_	W	Pder = (Tj - Ta)/θja (NOTE 2, 3)						

Legend: TBD = To Be Determined.

Note 1: IDD is current to run the chip alone without driving any load on the output pins.

2: TA = Ambient Temperature.

3: Maximum allowable power dissipation is the lower value of either the absolute maximum total power dissipation or derated power (PDER).

17.7 Timing Parameter Symbology

The timing parameter symbols have been created with one of the following formats:

- 1. TppS2ppS
- 2. TppS

Т			
F	Frequency	Т	Time
Lowerc	ase letters (pp) and their meanings:		
рр			
сс	CCP1	OSC	OSC1
ck	CLKOUT	rd	RD
CS	CS	rw	RD or WR
di	SDI	SC	SCK
do	SDO	SS	SS
dt	Data in	tO	ТОСКІ
io	I/O PORT	t1	T1CKI
mc	MCLR	wr	WR
Upperc	ase letters and their meanings:		
S			
F	Fall	Р	Period
Н	High	R	Rise
I	Invalid (High-impedance)	V	Valid
L	Low	Z	High-impedance

FIGURE 17-3: LOAD CONDITIONS



17.8 AC Characteristics: PIC16F882/883/884/886/887 (Industrial, Extended)



FIGURE 17-4: CLOCK TIMING

TABLE 17-1: CLOCK OSCILLATOR TIMING REQUIREMENTS -

Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +125^{\circ}C$										
Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions			
OS01	Fosc	External CLKIN Frequency ⁽¹⁾	DC	_	37	kHz	LP Oscillator mode			
			DC	—	4 MHz XT Oscillator mode		XT Oscillator mode			
			DC	—	20	MHz	HS Oscillator mode			
			DC	—	20	MHz	EC Oscillator mode			
		Oscillator Frequency ⁽¹⁾	—	32.768	—	kHz	LP Oscillator mode			
			0.1	—	4	MHz	XT Oscillator mode			
			1	—	20	MHz	HS Oscillator mode			
			DC	—	4	MHz	RC Oscillator mode			
OS02	Tosc	External CLKIN Period ⁽¹⁾	27	—	•	μs	LP Oscillator mode			
			250	—	•	ns	XT Oscillator mode			
			50	—	•	ns	HS Oscillator mode			
			50	—	٠	ns	EC Oscillator mode			
		Oscillator Period ⁽¹⁾	_	30.5	_	μs	LP Oscillator mode			
			250	—	10,000	ns	XT Oscillator mode			
			50	—	1,000	ns	HS Oscillator mode			
			250	—	_	ns	RC Oscillator mode			
OS03	Тсү	Instruction Cycle Time ⁽¹⁾	200	Тсү	DC	ns	TCY = 4/FOSC			
OS04*	TosH,	External CLKIN High,	2	_		μs	LP oscillator			
	TosL	External CLKIN Low	100	—	—	ns	XT oscillator			
			20	—	—	ns	HS oscillator			
OS05*	TosR,	External CLKIN Rise,	0	—	٠	ns	LP oscillator			
	TosF	External CLKIN Fall	0	—	•	ns	XT oscillator			
			0	—	•	ns	HS oscillator			

These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Instruction cycle period (Tcy) equals four times the input oscillator time base period. All specified values are Note 1: based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min" values with an external clock applied to OSC1 pin. When an external clock input is used, the "max" cycle time limit is "DC" (no clock) for all devices.

TABLE 17-2: OSCILLATOR PARAMETERS

-

Standard Operating	Standard Operating Conditions (unless otherwise stated)Operating Temperature $-40^{\circ}C \le T_A \le +125^{\circ}C$										
Param No.	Sym	Characteristic	Freq Tolerance	Min	Тур†	Max	Units	Conditions			
OS06	TWARM	Internal Oscillator Switch when running ⁽³⁾			_	2	Tosc	Slowest clock			
OS07	Tsc	Fail-Safe Sample Clock Period ⁽¹⁾	—		21	—	ms	LFINTOSC/64			
OS08	HFosc	Internal Calibrated	±1%	7.92	8.0	8.08	MHz	VDD = 3.5V, 25°C			
		HFINTOSC Frequency ⁽²⁾	±2%	7.84	8.0	8.16	MHz	$2.5V \le VDD \le 5.5V$, $0^{\circ}C \le TA \le +85^{\circ}C$			
			±5%	7.60	8.0	8.40	MHz	$\begin{array}{l} 2.0V \leq V \text{DD} \leq 5.5 \text{V}, \\ -40^{\circ}\text{C} \leq \text{TA} \leq +85^{\circ}\text{C} \text{ (Ind.)}, \\ -40^{\circ}\text{C} \leq \text{TA} \leq +125^{\circ}\text{C} \text{ (Ext.)} \end{array}$			
OS09*	LFosc	Internal Uncalibrated LFINTOSC Frequency		15	31	45	kHz				
OS10*	Tiosc	HFINTOSC Oscillator	_	5.5	12	24	μs	VDD = 2.0V, -40°C to +85°C			
	ST	Wake-up from Sleep	—	3.5	7	14	μs	VDD = 3.0V, -40°C to +85°C			
		Start-up Time	—	3	6	11	μs	VDD = 5.0V, -40°C to +85°C			

* These parameters are characterized but not tested.

Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only † and are not tested.

- Note 1: Instruction cycle period (TCY) equals four times the input oscillator time base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min" values with an external clock applied to the OSC1 pin. When an external clock input is used, the "max" cycle time limit is "DC" (no clock) for all devices.
 - 2: To ensure these oscillator frequency tolerances, VDD and Vss must be capacitively decoupled as close to the device as possible. 0.1 μ F and 0.01 μ F values in parallel are recommended.

3: By design.


FIGURE 17-5: CLKOUT AND I/O TIMING

TABLE 17-3: CLKOUT AND I/O TIMING PARAMETERS

Standar Operatin	Standard Operating Conditions (unless otherwise stated) Operating Temperature $-40^{\circ}C \le TA \le +125^{\circ}C$									
Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions			
OS11	TosH2ckL	Fosc↑ to CLKOUT↓ ⁽¹⁾	—	_	70	ns	VDD = 5.0V			
OS12	TosH2ckH	Fosc↑ to CLKOUT↑ ⁽¹⁾	—	—	72	ns	VDD = 5.0V			
OS13	TCKL2IOV	CLKOUT↓ to Port out valid ⁽¹⁾	—	—	20	ns				
OS14	ТюV2скН	Port input valid before CLKOUT↑ ⁽¹⁾	Tosc + 200 ns	—		ns				
OS15*	TosH2IoV	Fosc↑ (Q1 cycle) to Port out valid	—	50	70	ns	VDD = 5.0V			
OS16	TosH2ıol	Fosc [↑] (Q2 cycle) to Port input invalid (I/O in hold time)	50	—	_	ns	VDD = 5.0V			
OS17	TioV2osH	Port input valid to Fosc↑ (Q2 cycle) (I/O in setup time)	20	—	—	ns				
OS18	TIOR	Port output rise time ⁽²⁾		15 40	72 32	ns	VDD = 2.0V VDD = 5.0V			
OS19	TIOF	Port output fall time ⁽²⁾	—	28 15	55 30	ns	VDD = 2.0V VDD = 5.0V			
OS20*	TINP	INT pin input high or low time	25	—		ns				
OS21*	Trap	PORTA interrupt-on-change new input level time	Тсү	_	_	ns				

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated.

Note 1: Measurements are taken in RC mode where CLKOUT output is 4 x Tosc.

2: Includes OSC2 in CLKOUT mode.



FIGURE 17-6: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND POWER-UP TIMER TIMING





TABLE 17-4:RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER, POWER-UP TIMER
AND BROWN-OUT RESET PARAMETERS

Standard Operating Conditions (unless otherwise stated)Operating Temperature $-40^{\circ}C \le TA \le +125^{\circ}C$									
Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions		
30	TMCL	MCLR Pulse Width (low)	2 5		_	μs μs	VDD = 5V, -40°C to +85°C VDD = 5V		
31	Twdt	Watchdog Timer Time-out Period (No Prescaler)	10 10	16 16	29 31	ms ms	VDD = 5V, -40°C to +85°C VDD = 5V		
32	Tost	Oscillation Start-up Timer Period ^(1, 2)	—	1024	—	Tosc	(NOTE 3)		
33*	TPWRT	Power-up Timer Period	40	65	140	ms			
34*	Tioz	I/O High-impedance from MCLR Low or Watchdog Timer Reset	—	_	2.0	μs			
35	VBOR	Brown-out Reset Voltage	2.0	_	2.2	V	BOR4V bit = 0 (NOTE 4)		
35	VBOR	Brown-out Reset Voltage	TBD	4.0	TBD	V	BOR4V bit = 1 (NOTE 4)		
36*	VHYST	Brown-out Reset Hysteresis	—	50	—	mV			
37*	TBOR	Brown-out Reset Minimum Detection Period	100		—	μs	$VDD \leq VBOR$		

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

- 2: By design.
- 3: Period of the slower clock.
- **4:** To ensure these voltage tolerances, VDD and VSS must be capacitively decoupled as close to the device as possible. 0.1 μF and 0.01 μF values in parallel are recommended.

Note 1: Instruction cycle period (TcY) equals four times the input oscillator time base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min" values with an external clock applied to the OSC1 pin. When an external clock input is used, the "max" cycle time limit is "DC" (no clock) for all devices.

FIGURE 17-8: TIMER0 AND TIMER1 EXTERNAL CLOCK TIMINGS



TABLE 17-5: TIMER0 AND TIMER1 EXTERNAL CLOCK REQUIREMENTS

Operatir	ng Temperatur	re -40°C	\leq TA \leq +125°C	e stateu)					
Param No.	Sym		Characteristic		Min	Тур†	Max	Units	Conditions
40*	T⊤0H	T0CKI High F	Pulse Width	No Prescaler	0.5 TCY + 20	—	_	ns	
				With Prescaler	10	—		ns	
41*	TT0L	T0CKI Low F	ulse Width	No Prescaler	0.5 TCY + 20	—		ns	
		With Prescaler		10	—		ns		
42*	Тт0Р	T0CKI Period	1		Greater of: 20 or <u>Tcy + 40</u> N	_		ns	N = prescale value (2, 4,, 256)
45*	T⊤1H	T1CKI High Time	Synchronous, No Prescaler		0.5 TCY + 20			ns	
			Synchronous, with Prescaler		15	—	_	ns	
			Asynchronous		30	_	_	ns	
46*	TT1L	T1CKI Low Time	Synchronous, No Prescaler		0.5 TCY + 20	_		ns	
			Synchronous, with Prescaler		15		_	ns	
			Asynchronous		30	—	_	ns	
47*	47* T⊤1P T1CKI Input Synchronous Period			Greater of: 30 or <u>Tcy + 40</u> N	_	_	ns	N = prescale value (1, 2, 4, 8)	
			Asynchronous		60	—	_	ns	
48	FT1	Timer1 Oscill (oscillator en	ator Input Frequebled by setting	iency Range bit T1OSCEN)	—	32.768	—	kHz	
49*	TCKEZTMR1	Delay from E Increment	xternal Clock Ed	dge to Timer	2 Tosc	_	7 Tosc	_	Timers in Sync mode

Standard Operating Conditions (unless otherwise stated)

These parameters are characterized but not tested.

t Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.





TABLE 17-6: CAPTURE/COMPARE/PWM REQUIREMENTS (ECCP)

Standar Operatir	Standard Operating Conditions (unless otherwise stated)Operating Temperature $-40^{\circ}C \le TA \le +125^{\circ}C$									
Param No.	Sym	Character	istic	Min	Тур†	Max	Units	Conditions		
CC01*	TccL	CCP1 Input Low Time	No Prescaler	0.5Tcy + 20	_	_	ns			
			With Prescaler	20	_		ns			
CC02*	TccH	CCP1 Input High Time	No Prescaler	0.5TCY + 20	_	_	ns			
			With Prescaler	20	_	_	ns			
CC03*	TccP	CCP1 Input Period		<u>3Tcy + 40</u> N	—	_	ns	N = prescale value (1, 4 or 16)		
		parametera ara abaraata	rized but not tested					10)		

These parameters are characterized but not tested.

Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance † only and are not tested.

TABLE 17-7: COMPARATOR SPECIFICATIONS

Standard Operating Conditions (unless otherwise stated)

Operating Temperature -40°C ≤ TA ≤ +125°C										
Param No.	Sym	Characteristics		Min	Тур†	Мах	Units	Comments		
CM01	Vos	Input Offset Voltage		—	± 5.0	± 10	mV	(Vdd - 1.5)/2		
CM02	Vсм	Input Common Mode Voltage		0	_	Vdd - 1.5	V			
CM03*	CMRR	Common Mode Rejection Ratio		+55			dB			
CM04*	Trt	Response Time	Falling	_	150	600	ns	(NOTE 1)		
			Rising	—	200	1000	ns			
CM05*	Тмc2coV	Comparator Mode Change to Output Valid		_		10	μs			

These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Response time is measured with one comparator input at (VDD - 1.5)/2 - 100 mV to (VDD - 1.5)/2 + 20 mV.

COMPARATOR VOLTAGE REFERENCE (CVREF) SPECIFICATIONS TABLE 17-8:

Standard Operating Conditions (unless otherwise stated)

 $-40^{\circ}C < T_A < +125^{\circ}C$ Operating temperature

Param No.	Sym	Characteristics	Min	Тур†	Max	Units	Comments			
CV01*	CLSB	Step Size ⁽²⁾	_	Vdd/24 Vdd/32		V V	Low Range (VRR = 1) High Range (VRR = 0)			
CV02*	CACC	Absolute Accuracy		—	± 1/2 ± 1/2	LSb LSb	Low Range (VRR = 1) High Range (VRR = 0)			
CV03*	CR	Unit Resistor Value (R)		2k		Ω				
CV04*	CST	Settling Time ⁽¹⁾	_	—	10	μs				

These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Settling time measured while VRR = 1 and VR<3:0> transitions from '0000' to '1111'.

2: See Section 8.10 "Comparator Voltage Reference" for more information.

TABLE 17-9: PIC16F882/883/884/886/887 A/D CONVERTER (ADC) CHARACTERISTICS

Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +125^{\circ}C$									
Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions		
AD01	NR	Resolution	—	_	10 bits	bit			
AD02	EIL	Integral Error			±1	LSb	VREF = 5.12V		
AD03	Edl	Differential Error	-	_	±1	LSb	No missing codes to 10 bits VREF = 5.12V		
AD04	EOFF	Offset Error	—	1.5	TBD	LSb	VREF = 5.12V		
AD07	Egn	Gain Error	-	±1	TBD	LSb	VREF = 5.12V		
AD06 AD06A	Vref	Reference Voltage ⁽³⁾	2.2 2.7	_	— Vdd	V	Absolute minimum to ensure 1 LSb accuracy		
AD07	VAIN	Full-Scale Range	Vss	_	Vref	V			
AD08	Zain	Recommended Impedance of Analog Voltage Source	_	_	10	kΩ			
AD09*	IREF	VREF Input Current ⁽³⁾	10	_	1000	μA	During VAIN acquisition. Based on differential of VHOLD to VAIN.		
			_	_	50	μA	During A/D conversion cycle.		

Legend: TBD = To Be Determined.

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Total Absolute Error includes integral, differential, offset and gain errors.

- **2:** The A/D conversion result never decreases with an increase in the input voltage and has no missing codes.
- 3: ADC VREF is from external VREF or VDD pin, whichever is selected as reference input.
- 4: When ADC is off, it will not consume any current other than leakage current. The power-down current specification includes any such leakage from the ADC module.

TABLE 17-10: PIC16F882/883/884/886/887 A/D CONVERSION REQUIREMENTS

Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +125^{\circ}C$									
Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions		
AD130*	TAD	A/D Clock Period	1.6	—	9.0	μs	Tosc-based, VREF≥3.0V		
			3.0	—	9.0	μs	Tosc-based, VREF full range		
		A/D Internal RC					ADCS<1:0> = 11 (ADRC mode)		
		Oscillator Period	3.0	6.0	9.0	μs	At VDD = $2.5V$		
			1.6	4.0	6.0	μs	At VDD = 5.0V		
AD131	TCNV	Conversion Time (not including Acquisition Time) ⁽¹⁾	_	11		Tad	Set GO/DONE bit to new data in A/D Result register		
AD132*	TACQ	Acquisition Time		11.5	_	μs			
AD133*	TAMP	Amplifier Settling Time	_	—	5	μs			
AD134	Tgo	Q4 to A/D Clock Start	_	Tosc/2	_	—			
			_	Tosc/2 + Tcy			If the A/D clock source is selected as RC, a time of TcY is added before the A/D clock starts. This allows the SLEEP instruction to be executed.		

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: ADRESH and ADRESL registers may be read on the following TCY cycle.

2: See Section 9.3 "A/D Acquisition Requirements" for minimum conditions.









NOTES:

18.0 DC AND AC CHARACTERISTICS GRAPHS AND TABLES

Graphs are not available at this time.

NOTES:

19.0 PACKAGING INFORMATION

19.1 Package Marking Information

28-Lead PDIP





28-Lead QFN





Example



Example



Example



Legend:	XXX Y YY WW NNN @3 *	Customer-specific information Year code (last digit of calendar year) Year code (last 2 digits of calendar year) Week code (week of January 1 is week '01') Alphanumeric traceability code Pb-free JEDEC designator for Matte Tin (Sn) This package is Pb-free. The Pb-free JEDEC designator ((e3)) can be found on the outer packaging for this package.						
Note:	In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.							

19.1 Package Marking Information (Continued)



44-Lead QFN



44-Lead TQFP



Example



Example



Example



Legend	: XXX Y YY WW NNN @3 *	Customer-specific information Year code (last digit of calendar year) Year code (last 2 digits of calendar year) Week code (week of January 1 is week '01') Alphanumeric traceability code Pb-free JEDEC designator for Matte Tin (Sn) This package is Pb-free. The Pb-free JEDEC designator (e3) can be found on the outer packaging for this package.							
Note:	In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.								

19.2 Package Details

The following sections give the technical details of the packages.

28-Lead Skinny Plastic Dual In-Line (SP or PJ) – 300 mil Body [SPDIP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units			
Dimension	n Limits	MIN	NOM	MAX
Number of Pins	Ν		28	•
Pitch	е		.100 BSC	
Top to Seating Plane	Α	-	-	.200
Molded Package Thickness	A2	.120	.135	.150
Base to Seating Plane	A1	.015	-	-
Shoulder to Shoulder Width	E	.290	.310	.335
Molded Package Width	E1	.240	.285	.295
Overall Length	D	1.345	1.365	1.400
Tip to Seating Plane	L	.110	.130	.150
Lead Thickness	с	.008	.010	.015
Upper Lead Width	b1	.040	.050	.070
Lower Lead Width	b	.014	.018	.022
Overall Row Spacing §	eB	-	_	.430

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

2. § Significant Characteristic.

3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" per side.

4. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-070B

28-Lead Plastic Small Outline (SO or OI) – Wide, 7.50 mm Body [SOIC]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

2. § Significant Characteristic.

3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15 mm per side.

4. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-052B

28-Lead Plastic Shrink Small Outline (SS) – 5.30 mm Body [SSOP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units	MILLIMETERS			
Dimens	ion Limits	MIN	NOM	MAX	
Number of Pins	Ν		28		
Pitch	е		0.65 BSC		
Overall Height	А	-	-	2.00	
Molded Package Thickness	A2	1.65	1.75	1.85	
Standoff	A1	0.05	-	-	
Overall Width	E	7.40	7.80	8.20	
Molded Package Width	E1	5.00	5.30	5.60	
Overall Length	D	9.90	10.20	10.50	
Foot Length	L	0.55	0.75	0.95	
Footprint	L1	1.25 REF			
Lead Thickness	С	0.09	-	0.25	
Foot Angle	¢	0°	4°	8°	
Lead Width	b	0.22	_	0.38	

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

2. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.20 mm per side.

- 3. Dimensioning and tolerancing per ASME Y14.5M.
 - BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-073B

28-Lead Plastic Quad Flat, No Lead Package (MM) – 6x6x0.9 mm Body [QFN-S] with 0.40 mm Contact Length

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		MILLIMETERS		
Dimensio	on Limits	MIN	NOM	MAX
Number of Pins	Ν		28	
Pitch	е		0.65 BSC	
Overall Height	Α	0.80	0.90	1.00
Standoff	A1	0.00	0.02	0.05
Contact Thickness	A3		0.20 REF	
Overall Width	Е		6.00 BSC	
Exposed Pad Width	E2	3.65	3.70	4.70
Overall Length	D		6.00 BSC	
Exposed Pad Length	D2	3.65	3.70	4.70
Contact Width	b	0.23	0.38	0.43
Contact Length	L	0.30	0.40	0.50
Contact-to-Exposed Pad	К	0.20	-	-

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

2. Package is saw singulated.

3. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-124B

40-Lead Plastic Dual In-Line (P or PL) – 600 mil Body [PDIP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		INCHES		
Dimensior	n Limits	MIN	NOM	MAX
Number of Pins	Ν	40		
Pitch	е	.100 BSC		
Top to Seating Plane	Α	-	-	.250
Molded Package Thickness	A2	.125	-	.195
Base to Seating Plane	A1	.015	-	-
Shoulder to Shoulder Width	E	.590	-	.625
Molded Package Width	E1	.485	-	.580
Overall Length	D	1.980	-	2.095
Tip to Seating Plane	L	.115	-	.200
Lead Thickness	С	.008	-	.015
Upper Lead Width	b1	.030	-	.070
Lower Lead Width	b	.014	_	.023
Overall Row Spacing §	eB		_	.700

Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. § Significant Characteristic.
- 3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" per side.
- 4. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-016B

44-Lead Plastic Quad Flat, No Lead Package (ML) – 8x8 mm Body [QFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



TOP VIEW

BOTTOM VIEW



Units		MILLIMETERS		
Dimensio	n Limits	MIN	NOM	MAX
Number of Pins	Ν		44	
Pitch	е		0.65 BSC	
Overall Height	А	0.80	0.90	1.00
Standoff	A1	0.00	0.02	0.05
Contact Thickness	A3		0.20 REF	
Overall Width	E		8.00 BSC	
Exposed Pad Width	E2	6.30	6.45	6.80
Overall Length	D		8.00 BSC	
Exposed Pad Length	D2	6.30	6.45	6.80
Contact Width	b	0.25	0.30	0.38
Contact Length	L	0.30	0.40	0.50
Contact-to-Exposed Pad	К	0.20	-	-

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

2. Package is saw singulated.

- 3. Dimensioning and tolerancing per ASME Y14.5M.
 - BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-103B

44-Lead Plastic Thin Quad Flatpack (PT) – 10x10x1 mm Body, 2.00 mm Footprint [TQFP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units		MILLIMETERS		
D	imension Limits	MIN	NOM	MAX	
Number of Leads	N	44			
Lead Pitch	е	0.80 BSC			
Overall Height	А	-	-	1.20	
Molded Package Thickness	A2	0.95	1.00	1.05	
Standoff	A1	0.05	-	0.15	
Foot Length	L	0.45	0.60	0.75	
Footprint	L1	1.00 REF			
Foot Angle	φ	0°	3.5°	7°	
Overall Width	E	12.00 BSC			
Overall Length	D	12.00 BSC			
Molded Package Width	E1	10.00 BSC			
Molded Package Length	D1	10.00 BSC			
Lead Thickness	С	0.09	-	0.20	
Lead Width	b	0.30	0.37	0.45	
Mold Draft Angle Top	α	11°	12°	13°	
Mold Draft Angle Bottom	β	11°	12°	13°	

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

2. Chamfers at corners are optional; size may vary.

3. Dimensions D1 and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.25 mm per side.

4. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-076B

NOTES:

APPENDIX A: DATA SHEET REVISION HISTORY

Revision A (5/2006)

Initial release of this data sheet.

Revision B (7/2006)

Pin Diagrams (44-Pin QFN drawing); Revised Table 2-1, Addr. 1DH (CCP2CON); Section 3.0, 3.1; Section 3.4.4.6; Table 3; Table 3-1 (ANSEL); Table 3-3 (CCP2CON); Register 3-1; Register 3.2; Register 3-3; Register 3-4; Register 3-9; Register 3-10; Register 3-11; Register 3-12; Register 3-14; Table 3-5 (ANSEL); Figure 3-5; Figure 3-11; Figure 8-2; Figure 8-3; Figure 9-1; Register 9-1; Section 9.1.4; Example 10-4; Figure 11-5; Table 11-5 (P1M); Section 11.5.2; Section 11.5.7, Number 4; Table 11-7 (CCP2CON); Section 12.3.1 (Para. 3); Figure 12-6 (Title); Sections 14.2, 14.3 and 14.4 DC Characteristics (Max); Table 14-4 (OSCCON); Section 14.3 (TMR0); Section 14.3.2 (TMR0).

Revision C

Section 19.0 Packaging Information: Replaced package drawings and added note. Added PIC16F882 part number. Replaced PICmicro with PIC.

Revision D

Replaced Package Drawings (Rev. AM); Replaced Development Support Section; Revised Product ID Section.

APPENDIX B: MIGRATING FROM OTHER PIC[®] DEVICES

This discusses some of the issues in migrating from other PIC devices to the PIC16F88X Family of devices.

B.1 PIC16F87X to PIC16F88X

TABLE B-1: FEATURE COMPARISON

Feature	PIC16F87X	PIC16F88X
Max Operating Speed	20 MHz	20 MHz
Max Program Memory (Words)	8192	8192
SRAM (bytes)	368	368
A/D Resolution	10-bit	10-bit
Data EEPROM (Bytes)	256	256
Timers (8/16-bit)	2/1	2/1
Oscillator Modes	4	8
Brown-out Reset	Y	Y (2.1V/4V)
Software Control Option of WDT/BOR	Ν	Y
Internal Pull-ups	RB<7:4>	RB<7:0>, MCLR
Interrupt-on-change	RB<7:4>	RB<7:0>
Comparator	2	2
References	CVREF	CVREF and VP6
ECCP/CCP	0/2	1/1
Ultra Low-Power Wake-Up	Ν	Y
Extended WDT	N	Y
INTOSC Frequencies	N	32 kHz-8 MHz
Clock Switching	N	Y
MSSP	Standard	w/Slave Address Mask
USART	AUSART	EUSART
ADC Channels	8	14

Note: This device has been designed to perform to the parameters of its data sheet. It has been tested to an electrical specification designed to determine its conformance with these parameters. Due to process differences in the manufacture of this device, this device may have different performance characteristics than its earlier version. These differences may cause this device to perform differently in your application than the earlier version of this device. NOTES:

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Device:	PIC16F883F ⁽¹⁾ , PIC16F883FT ^(1, 2) , PIC16F884F ⁽¹⁾ PIC16F884FT ^(1, 2) , PIC16F886F ⁽¹⁾ , PIC16F886FT ^(1, 2) , PIC16F887F ⁽¹⁾ , PIC16F887FT ^(1, 2) VDD range 2.0V to 5.5V	
Temperature Range:	I = -40° C to $+85^{\circ}$ C (Industrial) E = -40° C to $+125^{\circ}$ C (Extended)	
Package:	ML = Quad Flat No Leads (QFN) P = Plastic DIP PT = Plastic Thin-Quad Flatpack (TQFP) SO = Plastic Small Outline (SOIC) (7.50 mm) SS = Plastic Shrink Small Outline	 Note 1: F = Standard Voltage Range 2: T = in tape and reel SSOP, SOIC and QFN packages only.
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