

### **DUAL ULTRA MICROPOWER RAIL-TO-RAIL CMOS OPERATIONAL AMPLIFIER**

#### **GENERAL DESCRIPTION**

The ALD2706A/ALD2706B/ALD2706 is a dual monolithic CMOS micropower high slew-rate operational amplifier intended for a broad range of analog applications using  $\pm 1V$  to  $\pm 5V$  dual power supply systems, as well as +2V to +10V battery operated systems. All device characteristics are specified for +5V single supply or  $\pm 2.5V$  dual supply systems. Supply current is  $80\mu A$  maximum at 5V supply voltage. It is manufactured with Advanced Linear Devices' enhanced A CMOS silicon gate CMOS process.

The ALD2706A/ALD2706B/ALD2706 is designed to offer a trade-off of performance parameters providing a wide range of desired specifications. It has been developed specifically for the +5V single supply or  $\pm 1V$  to  $\pm 5V$  dual supply user and offers the popular industry standard pin configuration.

Several important characteristics of the device make application easier to implement at those voltages. First, each operational amplifier can operate with rail to rail input and output voltages. This means the signal input voltage and output voltage can be equal to the positive and negative supply voltages. This feature allows numerous analog serial stages and flexibility in input signal bias levels. Second, each device was designed to accommodate mixed applications where digital and analog circuits may operate off the same power supply or battery. Third, the output stage can typically drive up to 25pF capacitive and  $20 \mathrm{K}\Omega$  resistive loads. These features, combined with extremely low input currents, high open loop voltage gain of  $100 \mathrm{V/mV}$ , useful bandwidth of  $200 \mathrm{KHz}$ , a slew rate of  $0.1 \mathrm{V/\mu s}$ , low offset voltage and temperature drift, make the ALD2706A/ALD2706B/ALD2706 a versatile, micropower dual operational amplifier.

A typical ALD2706A/ALD2706B/ALD2706 has the capacity to process a 0.998V amplitude analog signal with only 1.000V single supply voltage, while requiring only 0.1pA input bias current. Additionally, robust design and rigorous screening make this device especially suitable for operation in temperature-extreme environments and rugged conditions.

#### **FEATURES**

- Typical 20μA supply current per amplifier
- All parameters specified for +5V single supply or ±2.5V dual supply systems
- Rail-to-rail input and output voltage ranges
- · Unity gain stable
- Extremely low input bias currents -- 0.1pA
- High source impedance applications
- Dual power supply ±1.0V to ±5.0V
- Single power supply +2V to +10V
- High voltage gain
- Unity gain bandwidth of 0.2MHz
- Slew rate of 0.1V/μs
- Symmetrical output drive
- Suitable for rugged, temperature-extreme environments

#### **APPLICATIONS**

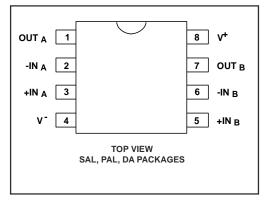
- Voltage follower/buffer/amplifier
- · Charge integrator
- · Photodiode amplifier
- Data acquisition systems
- High performance portable instruments
- Signal conditioning circuits
- · Sensor and transducer amplifiers
- Low leakage amplifiers
- Active filters
- · Sample/Hold amplifier
- Picoammeter
- · Current to voltage converter

### **ORDERING INFORMATION** ("L" suffix denotes lead-free (RoHS))

Operat	Operating Temperature Range										
0°C to +70°C	0°C to +70°C	-55°C to 125°C									
8-Pin	8-Pin	8-Pin									
Small Outline	Plastic Dip	CERDIP									
Package (SOIC)	Package	Package									
ALD2706ASAL	ALD2706APAL	ALD2706ADA									
ALD2706BSAL	ALD2706BPAL	ALD2706BDA									
ALD2706SAL	ALD2706PAL	ALD2706DA									

<sup>\*</sup> Contact factory for leaded (non-RoHS) or high temperature versions.

### **PIN CONFIGURATION**



### **ABSOLUTE MAXIMUM RATINGS**

Supply voltage, V+	10.6V
Differential input voltage range	-0.3V to V++0.3V
Power dissipation	600 mW
Operating temperature range SAL, PAL packages	0°C to +70°C
DA package	55°C to +125°C
Storage temperature range	65°C to +150°C
Lead temperature, 10 seconds	+260°C

**CAUTION:** ESD Sensitive Device. Use static control procedures in ESD controlled environment.

# OPERATING ELECTRICAL CHARACTERISTICS $T_A = 25^{\circ}C$ $V_S = \pm 2.5V$ unless otherwise specified

			2706A			2706B			2706			Test
Parameter	Symbol	Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Unit	Conditions
Supply Voltage	V <sub>S</sub> V+	±1.0 2.0		±5.0 10.0	±1.0 2.0		±5.0 10.0	±1.0 2.0		±5.0 10.0	V V	Dual Supply Single Supply
Input Offset Voltage	Vos			2.0 2.8			5.0 5.8			10.0 11.0	mV mV	$R_S \le 100K\Omega$ $0^{\circ}C \le T_A \le +70^{\circ}C$
Input Offset Current	I <sub>OS</sub>		0.1	20 200		0.1	20 200		0.1	20 200	pA pA	$T_A = 25^{\circ}C$ $0^{\circ}C \le T_A \le +70^{\circ}C$
Input Bias Current	IB		0.1	20 200		0.1	20 200		0.1	20 200	pA pA	$T_A = 25^{\circ}C$ $0^{\circ}C \le T_A \le +70^{\circ}C$
Input Voltage Range	V <sub>IR</sub>	-0.3 -2.8		5.3 2.8	-0.3 -2.8		5.3 2.8	-0.3 -2.8		5.3 2.8	V V	V+ = +5 V <sub>S</sub> = ±2.5V
Input Resistance	R <sub>IN</sub>		10 <sup>12</sup>			10 <sup>12</sup>			10 <sup>12</sup>		Ω	
Input Offset Voltage Drift	TCVOS		7			7			10		μV/°C	R <sub>S</sub> ≤ 100KΩ
Power Supply Rejection Ratio	PSRR	65 65	80 80		65 65	80 80		60 60	80 80		dB dB	R <sub>S</sub> ≤ 100KΩ 0°C ≤ T <sub>A</sub> ≤ +70°C
Common Mode Rejection Ratio	CMRR	65 65	83 83		65 65	83 83		60 60	83 83		dB dB	$R_S \le 100 K\Omega$ $0^{\circ}C \le T_A \le +70^{\circ}C$
Large Signal Voltage Gain	A <sub>V</sub>	10 10	100 300		10 10	100 300		5 5	80 300		V/mV V/mV V/mV	$\begin{aligned} R_L &= 100 K \Omega \\ R_L &\geq 1 M \Omega \\ R_L &= 100 K \Omega \\ 0^{\circ} C &\leq T_A \leq +70^{\circ} C \end{aligned}$
Output Voltage	V <sub>O</sub> low V <sub>O</sub> high	4.99	0.001 4.999	0.01	4.99	0.001 4.999	0.01	4.99	0.001 4.999	0.01	V V	$R_L = 1M\Omega V^+ = +5V$ $0^{\circ}C \le T_A \le +70^{\circ}C$
Range	V <sub>O</sub> low V <sub>O</sub> high	2.25	-2.40 2.40	-2.25	2.25	-2.40 2.40	-2.25	2.25	-2.40 2.40	-2.25	V V	$R_L = 100K\Omega$ $0^{\circ}C \le T_A \le +70^{\circ}C$
Output Short Circuit Current	I <sub>SC</sub>		200			200			200		μΑ	
Supply Current	IS		50	80		50	80		50	80	μΑ	V <sub>IN</sub> =0V No Load
Power Dissipation	P <sub>D</sub>			400			400			400	μW	Both amplifiers V <sub>S</sub> = ±2.5V

### **OPERATING ELECTRICAL CHARACTERISTICS (cont'd)**

 $T_A = 25^{\circ}C$   $V_S = \pm 2.5V$  unless otherwise specified

			2706A			2706B			2706			Test
Parameter	Symbol	Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Unit	Conditions
Input Capacitance	C <sub>IN</sub>		1			1			1		pF	
Bandwidth	B <sub>W</sub>		200			200			200		KHz	
Slew Rate	S <sub>R</sub>		0.1			0.1			0.1		V/μs	$A_V = +1$ $R_L = 100$ K $\Omega$
Rise time	t <sub>r</sub>		1.0			1.0			1.0		μS	R <sub>L</sub> = 100KΩ
Overshoot Factor			20			20			20		%	$R_L = 100K\Omega$ $C_L = 25pF$
Settling Time	t <sub>s</sub>		10.0			10.0			10.0		μS	0.1% A <sub>V</sub> = -1 C <sub>L</sub> = 25pF R <sub>L</sub> = 100KΩ
Channel Separation	Cs		140			140			140		dB	A <sub>V</sub> = 100

 $T_A = 25^{\circ}C$   $V_S = \pm 1.0V$  unless otherwise specified

			2706A			2706B			2706			Test
Parameter	Symbol	Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Unit	Conditions
Power Supply Rejection Ratio	PSRR		80			80			80		dB	R <sub>S</sub> ≤ 1MΩ
Common Mode Rejection Ratio	CMRR		80			80			80		dB	R <sub>S</sub> ≤1MΩ
Large Signal Voltage Gain	A <sub>V</sub>		50			50			50		V/mV	$R_L = 1M\Omega$
Output Voltage Range	V <sub>O</sub> low V <sub>O</sub> high	0.9	-0.95 0.95	-0.9	0.9	95 0.95	-0.9	0.9	-0.95 0.95	-0.9	V V	$R_L = 1M\Omega$
Bandwidth	B <sub>W</sub>		0.2			0.2			0.2		MHz	
Slew Rate	S <sub>R</sub>		0.1			0.1			0.1		V/μs	A <sub>V</sub> =+1 C <sub>L =</sub> 25pF

 $V_S = \pm 2.5 V -55^{\circ}C \le T_A \le +125^{\circ}C$  unless otherwise specified

		27	706ADA		:	2706BD	Α	2	2706DA			Test
Parameter	Symbol	Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Unit	Conditions
Input Offset Voltage	Vos			3.0			6.0			12.0	mV	R <sub>S</sub> ≤ 100KΩ
Input Offset Current	I <sub>OS</sub>			4.0			4.0			4.0	nA	
Input Bias Current	I <sub>B</sub>			4.0			4.0			4.0	nA	
Power Supply Rejection Ratio	PSRR	60	75		60	75		60	75		dB	R <sub>S</sub> ≤ 1MΩ
Common Mode Rejection Ratio	CMRR	60	83		60	83		60	83		dB	R <sub>S</sub> ≤ 1MΩ
Large Signal Voltage Gain	A <sub>V</sub>	10	50		10	50		5	50		V/mV	$R_L = 1M\Omega$
Output Voltage Range	V <sub>O</sub> low V <sub>O</sub> high	2.25	-2.40 2.40	-2.25	2.25	-2.40 2.40	-2.25	2.25	-2.40 2.40	-2.25	V V	$R_L = 1M\Omega$

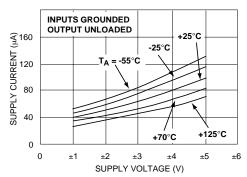
#### **Design & Operating Notes:**

- 1. The ALD2706A/ALD2706B/ALD2706 CMOS operational amplifier uses a 3 gain stage architecture and an improved frequency compensation scheme to achieve large voltage gain, high output driving capability, and better frequency stability. In a conventional CMOS operational amplifier design, compensation is achieved with a pole splitting capacitor together with a nulling resistor. This method is, however, very bias dependent and thus cannot accommodate the large range of supply voltage operation as is required from a stand alone CMOS operational amplifier. The ALD2706A/ALD2706B/ALD2706 is internally compensated for unity gain stability using a novel scheme that does not use a nulling resistor. This scheme produces a clean single pole roll off in the gain characteristics while providing for more than 70 degrees of phase margin at the unity gain frequency.
- 2. The ALD2706A/ALD2706B/ALD2706 has complementary p-channel and n-channel input differential stages connected in parallel to accomplish rail to rail input common mode voltage range. This means that with the ranges of common mode input voltage close to the power supplies, one of the two differential stages is switched off internally. To maintain compatibility with other operational amplifiers, this switching point has been selected to be about 1.5V below the positive supply voltage. Since offset voltage trimming on the ALD2706A/ALD2706B/ALD2706 is made when the input voltage is symmetrical to the supply voltages, this internal switching does not affect a large variety of applications such as an inverting amplifier or non-inverting amplifier with a gain larger than 2.5 (5V operation), where the common mode voltage does not make excursions above this switching point. The user should however, be aware that this switching does take place if the operational amplifier is connected as a unity gain buffer and should make provision in his design to allow for input offset voltage variations.
- 3. The input bias and offset currents are essentially input protection diode reverse bias leakage currents, and are typically less than

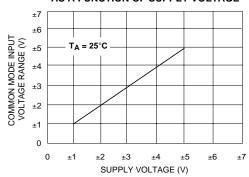
- 1pA at room temperature. This low input bias current assures that the analog signal from the source will not be distorted by input bias currents. Normally, this extremely high input impedance of greater than  $10^{12}\Omega$  would not be a problem as the source impedance would limit the node impedance. However, for applications where source impedance is very high, it may be necessary to limit noise and hum pickup through proper shielding.
- 4. The output stage consists of class AB complementary output drivers, capable of driving a low resistance load. The output voltage swing is limited by the drain to source on-resistance of the output transistors as determined by the bias circuitry, and the value of the load resistor. When connected in the voltage follower configuration, the oscillation resistant feature, combined with the rail to rail input and output feature, makes an effective analog signal buffer for medium to high source impedance sensors, transducers, and other circuit networks.
- 5. The ALD2706A/ALD2706B/ALD2706 operational amplifier has been designed to provide full static discharge protection. Internally, the design has been carefully implemented to minimize latch up. However, care must be exercised when handling the device to avoid strong static fields that may degrade a diode junction, causing increased input leakage currents. In using the operational amplifier, the user is advised to power up the circuit before, or simultaneously with, any input voltages applied and to limit input voltages to not exceed 0.3V of the power supply voltage levels.
- 6. The ALD2706A/ALD2706B/ALD2706, with its micropower operation, offers numerous benefits in reduced power supply requirements, less noise coupling and current spikes, less thermally induced drift, better overall reliability due to lower self heating, and lower input bias current. It requires practically no warm up time as the chip junction heats up to only 0.1°C above ambient temperature under most operating conditions.

### TYPICAL PERFORMANCE CHARACTERISTICS

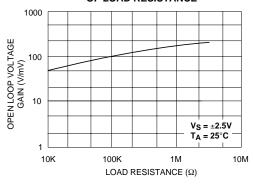
### SUPPLY CURRENT AS A FUNCTION OF SUPPLY VOLTAGE



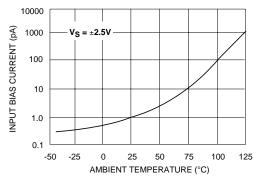
#### COMMON MODE INPUT VOLTAGE RANGE AS A FUNCTION OF SUPPLY VOLTAGE



### OPEN LOOP VOLTAGE GAIN AS AFUNCTION OF LOAD RESISTANCE

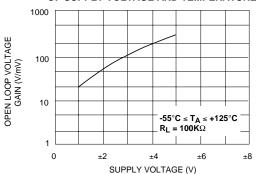


### INPUT BIAS CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE

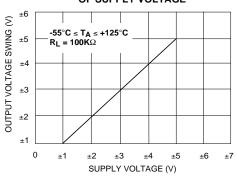


### TYPICAL PERFORMANCE CHARACTERISTICS (cont'd)

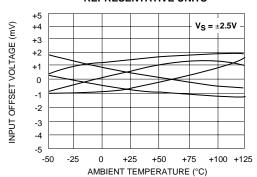
### OPEN LOOP VOLTAGE GAIN AS A FUNCTION OF SUPPLY VOLTAGE AND TEMPERATURE



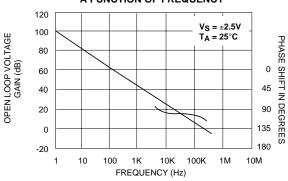
### OUTPUT VOLTAGE SWING AS A FUNCTION OF SUPPLY VOLTAGE



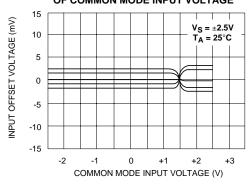
# INPUT OFFSET VOLTAGE AS A FUNCTION OF AMBIENT TEMPERATURE REPRESENTATIVE UNITS



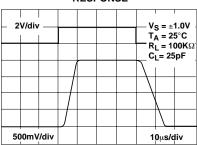
### OPEN LOOP VOLTAGE GAIN AS A FUNCTION OF FREQUENCY



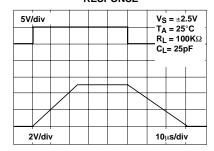
### INPUT OFFSET VOLTAGE AS A FUNCTION OF COMMON MODE INPUT VOLTAGE



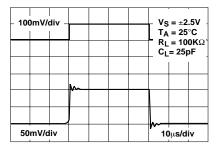
LARGE - SIGNAL TRANSIENT RESPONSE



### LARGE - SIGNAL TRANSIENT RESPONSE



SMALL - SIGNAL TRANSIENT RESPONSE

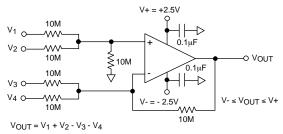


### **TYPICAL APPLICATIONS**

#### **RAIL-TO-RAIL VOLTAGE FOLLOWER/BUFFER**

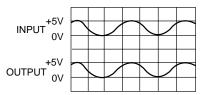
### 

# HIGH INPUT IMPEDANCE RAIL-TO-RAIL PRECISION DC SUMMING AMPLIFIER



 $R_{IN}$  = 10M $\Omega$  Accuracy limited by resistor tolerances and input offset voltage

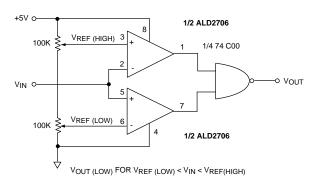
### **RAIL-TO-RAIL WAVEFORM**



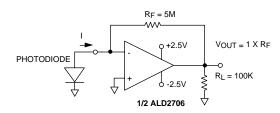
#### Performance waveforms.

Upper trace is the output of a Wien Bridge Oscillator. Lower trace is the output of Rail-to-Rail voltage follower.

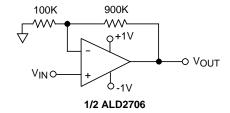
#### **RAIL-TO-RAIL WINDOW COMPARATOR**



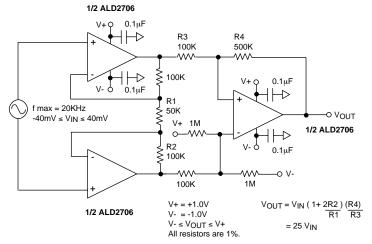
## PHOTO DETECTOR CURRENT TO VOLTAGE CONVERTER



### HIGH IMPEDANCE NON-INVERTING AMPLIFIER

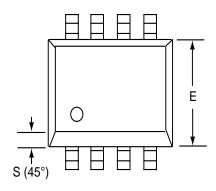


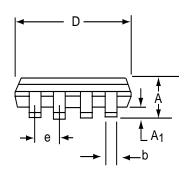
#### LOW VOLTAGE INSTRUMENTATION AMPLIFIER



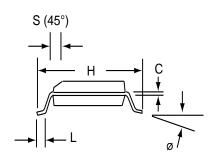
### **SOIC-8 PACKAGE DRAWING**

### 8 Pin Plastic SOIC Package



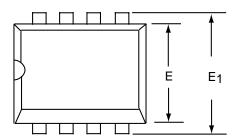


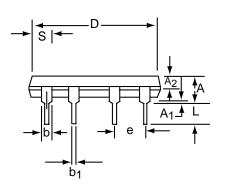
	Millim	eters	Inches				
Dim	Min	Max	Min	Max			
Α	1.35	1.75	0.053	0.069			
A <sub>1</sub>	0.10	0.25	0.004	0.010			
b	0.35	0.45	0.014	0.018			
С	0.18	0.25	0.007	0.010			
D-8	4.69	5.00	0.185	0.196			
Е	3.50	4.05	0.140	0.160			
е	1.27	BSC	0.050 BSC				
н	5.70	6.30	0.224	0.248			
L	0.60	0.937	0.024	0.037			
Ø	0°	8°	0°	8°			
s	0.25	0.50	0.010	0.020			



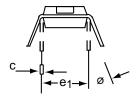
### **PDIP-8 PACKAGE DRAWING**

### 8 Pin Plastic DIP Package



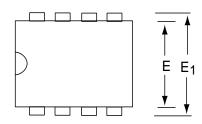


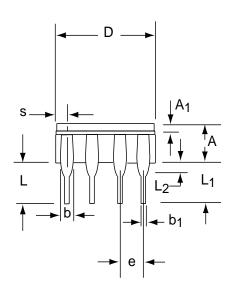
	Millim	neters	Inches				
Dim	Min	Max	Min	Max			
Α	3.81	5.08	0.105	0.200			
A <sub>1</sub>	0.38	1.27	0.015	0.050			
A <sub>2</sub>	1.27	2.03	0.050	0.080			
b	0.89	1.65	0.035	0.065			
b <sub>1</sub>	0.38	0.51	0.015	0.020			
С	0.20	0.30	0.008	0.012			
D-8	9.40	11.68	0.370	0.460			
E	5.59	7.11	0.220	0.280			
E <sub>1</sub>	7.62	8.26	0.300	0.325			
е	2.29	2.79	0.090	0.110			
e <sub>1</sub>	7.37	7.87	0.290	0.310			
L	2.79	3.81	0.110	0.150			
S-8	1.02	2.03	0.040	0.080			
Ø	0°	15°	0°	15°			

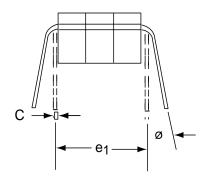


### **CERDIP-8 PACKAGE DRAWING**

### 8 Pin CERDIP Package







	Millin	neters	Inc	hes		
Dim	Min	Max	Min	Max		
Α	3.55	5.08	0.140	0.200		
A <sub>1</sub>	1.27	2.16	0.050	0.085		
b	0.97	1.65	0.038	0.065		
b <sub>1</sub>	0.36	0.58	0.014	0.023		
С	0.20	0.38	0.008	0.015		
D-8		10.29		0.405		
E	5.59	7.87	0.220	0.310		
E <sub>1</sub>	7.73	8.26	0.290	0.325		
е	2.54 E	BSC	0.100 BSC			
e <sub>1</sub>	7.62 E	BSC	0.300	0 BSC		
L	3.81	5.08	0.150	0.200		
L <sub>1</sub>	3.18		0.125			
L <sub>2</sub>	0.38	1.78	0.015	0.070		
S		2.49		0.098		
Ø	0°	15°	0°	15°		

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