ADVANCED
LINEAR
DEVICES, INC.

ALD4706A/ALD4706B
ALD4706

DUAL ULTRA MICROPOWER RAIL-TO-RAIL CMOS OPERATIONAL AMPLIFIER

GENERAL DESCRIPTION

The ALD4706A/ALD4706B/ALD4706 is a quad monolithic CMOS ultra micropower high slew-rate operational amplifier intended for a broad range of analog applications using ±1V to ±5V dual power supply systems, as well as +2V to +10V battery operated systems. All device characteristics are specified for +5V single supply or ±2.5V dual supply systems. Total supply current for four operational amplifiers is 200μA maximum at 5V supply voltage. It is manufactured with Advanced Linear Devices' enhanced ACMOS silicon gate CMOS process.

The ALD4706A/ALD4706B/ALD4706 is designed to offer a trade-off of performance parameters providing a wide range of desired specifications. It has been developed specifically for the +5V single supply or ±1V to ±5V dual supply user and offers the popular industry standard pin configuration of LM324 types and ICL7641 types.

Several important characteristics of the device make application easier to implement at these voltages. First, each operational amplifier can operate with rail-to-rail input and output voltages. This means the signal input voltage and output voltage can be equal to or near to the positive and negative supply voltages. This feature allows numerous analog serial stages and flexibility in input signal bias levels. Second, each device was designed to accommodate mixed applications where digital and analog circuits may operate off the same power supply or battery. Third, the output stage can typically drive up to 25pF capacitive and 20KΩ resistive loads. These features, combined with extremely low input currents, high open loop voltage gain of 100V/mV, useful bandwidth of 200KHz, a slew rate of 0.17V/μs, low power dissipation of 0.5mW, low offset voltage and temperature drift, make the ALD4706A/ALD4706B/ALD4706 a versatile, ultra micropower quad operational amplifier.

The ALD4706A/ALD4706B/ALD4706, designed and fabricated with silicon gate CMOS technology, offers 0.1pA typical input bias current. Due to low voltage and low power operation, reliability and operating characteristics, such as input bias currents and warm up time, are greatly improved. Additionally, robust design and rigorous screening make this device especially suitable for operation in temperature-extreme environments and rugged conditions.
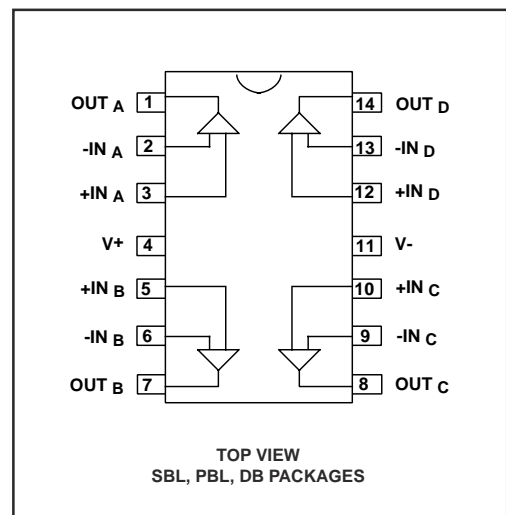
FEATURES

- All parameters specified for + 5V single supply or ± 2.5V dual supply systems
- Rail- to- rail input and output voltage ranges
- Unity gain stable
- Extremely low input bias currents -- 0.1pA
- High source impedance applications
- Dual power supply ±1.0V to ±5.0V
- Single power supply +2V to +10V
- High voltage gain
- Output short circuit protected
- Unity gain bandwidth of 0.2MHz
- Slew rate of 0.17V/μs
- Power dissipation of 20μA per op amp
- Symmetrical output drive
- Suitable for rugged, temperature-extreme environments

APPLICATIONS

- Voltage follower/buffer/amplifier
- Charge integrator
- Photodiode amplifier
- Data acquisition systems
- High performance portable instruments
- Signal conditioning circuits
- Sensor and transducer amplifiers
- Low leakage amplifiers
- Active filters
- Sample/Hold amplifier
- Picoammeter
- Current to voltage converter

ORDERING INFORMATION ("L" suffix denotes lead-free (RoHS))

Operating Temperature Range		
0°C to +70°C	0°C to +70°C	-55°C to 125°C
14-Pin Small Outline Package (SOIC)	14-Pin Plastic Dip Package	14-Pin CERDIP Package
ALD4706ASBL ALD4706BSBL ALD4706SBL	ALD4706APBL ALD4706BPBL ALD4706PBL	ALD4706ADB ALD4706BDB ALD4706DB

* Contact factory for leaded (non-RoHS) or high temperature versions.

PIN CONFIGURATION

OUT A	1	14	OUT D
-IN A	2	13	-IN D
+IN A	3	12	+IN D
V+	4	11	V-
+IN B	5	10	+IN C
-IN B	6	9	-IN C
OUT B	7	8	OUT C

TOP VIEW
SBL, PBL, DB PACKAGES

ABSOLUTE MAXIMUM RATINGS

Supply voltage, V+ —————————————————————————— 10.6V
Differential input voltage range ————————————————— -0.3V to V+ +0.3V
Power dissipation ——————————————————————————— 600 mW
Operating temperature range   SBL, PBL packages ———————————— 0°C to +70°C
                              DB package ———————————————— -55°C to +125°C
Storage temperature range ———————————————————— -65°C to +150°C
Lead temperature, 10 seconds ———————————————————— +260°C
CAUTION: ESD Sensitive Device. Use static control procedures in ESD controlled environment.

OPERATING ELECTRICAL CHARACTERISTICS
$T_A = 25°C$ $V_S = ±2.5V$ unless otherwise specified

Parameter	Symbol	4706A Min	Typ	Max	4706B Min	Typ	Max	4706 Min	Typ	Max	Unit	Test Conditions
Supply Voltage	V_S $V+$	±1.0 2.0		±5.0 10.0	±1.0 2.0		±5.0 10.0	±1.0 2.0		±5.0 10.0	V V	Dual Supply Single Supply
Input Offset Voltage	V_{OS}			2.0 2.8			5.0 5.8			10.0 11.0	mV mV	$R_S ≤ 100KΩ$ $0°C ≤ T_A ≤ +70°C$
Input Offset Current	I_{OS}		0.1	20 200		0.1	20 200		0.1	20 200	pA pA	$T_A = 25°C$ $0°C ≤ T_A ≤ +70°C$
Input Bias Current	I_B		0.1	20 200		0.1	20 200		0.1	20 200	pA pA	$T_A = 25°C$ $0°C ≤ T_A ≤ +70°C$
Input Voltage Range	V_{IR}	-0.3 -2.8		5.3 2.8	-0.3 -2.8		5.3 2.8	-0.3 -2.8		5.3 2.8	V V	$V+ = +5$ $V_S = ±2.5V$
Input Resistance	R_{IN}		10^{13}			10^{13}			10^{13}		Ω	
Input Offset Voltage Drift	TCV_{OS}		7			7			10		µV/°C	$R_S ≤ 100KΩ$
Power Supply Rejection Ratio	PSRR	65 65	83 83		65 65	83 83		60 60	83 83		dB dB	$R_S ≤ 100KΩ$ $0°C ≤ T_A ≤ +70°C$
Common Mode Rejection Ratio	CMRR	65 65	83 83		65 65	83 83		60 60	83 83		dB dB	$R_S ≤ 100KΩ$ $0°C ≤ T_A ≤ +70°C$
Large Signal Voltage Gain	A_V	10 10	60 300		10 10	60 300		7 7	50 300		V/mV V/mV V/mV	$R_L = 100KΩ$ $R_L ≥ 1MΩ$ $R_L = 100KΩ$
Output Voltage Range	V_O low V_O high	4.99	0.001 4.999	0.01	4.99	0.001 4.999	0.01	4.99	0.001 4.999	0.01	V V	$R_L = 1MΩ$ $V+ = +5V$ $0°C ≤ T_A ≤ +70°C$
	V_O low V_O high	2.25	-2.40 2.40	-2.25	2.25	-2.40 2.40	-2.25	2.25	-2.40 2.40	-2.25	V V	$R_L = 100KΩ$ $V_S = ±2.5V$ $0°C ≤ T_A ≤ +70°C$
Output Short Circuit Current	I_{SC}		200			200			200		µA	
Supply Current	I_S		120	200		120	200		120	200	µA	$V_{IN}=0V$ No Load
Power Dissipation	P_D			1.0			1.0			1.0	µW	All amplifiers $V_S = ±2.5V$

OPERATING ELECTRICAL CHARACTERISTICS (cont'd)

$T_A = 25°C$ $V_S = ±2.5V$ unless otherwise specified

Parameter	Symbol	4706A			4706B			4706			Unit	Test Conditions
		Min	Typ	Max	Min	Typ	Max	Min	Typ	Max		
Input Capacitance	C_{IN}		1			1			1		pF	
Bandwidth	B_W		200			200			200		KHz	
Slew Rate	S_R		0.17			0.17			0.17		V/µs	$R_L = 100KΩ$ $A_V = +1$
Rise time	t_r		1.0			1.0			1.0		µs	$R_L = 100KΩ$
Overshoot Factor			20			20			20		%	$R_L = 100KΩ$ $C_L = 25pF$
Settling Time	t_s		10.0			10.0			10.0		µs	0.1% $A_V = 1$ $C_L = 25pF$ $R_L = 100KΩ$
Channel Separation	C_S		140			140			140		dB	$A_V = 100$

$T_A = 25°C$ $V_S = ±1.0V$ unless otherwise specified

Parameter	Symbol	4706A			4706B			4706			Unit	Test Conditions
		Min	Typ	Max	Min	Typ	Max	Min	Typ	Max		
Power Supply Rejection Ratio	PSRR		80			80			80		dB	$R_S ≤ 1MΩ$
Common Mode Rejection Ratio	CMRR		80			80			80		dB	$R_S ≤ 1MΩ$
Large Signal Voltage Gain	A_V		50			50			50		V/mV	$R_L = 1MΩ$
Output Voltage Range	V_O low V_O high	.90	-0.95 0.95	-.90	.90	-0.95 0.95	-.90	.90	-0.95 0.95	-.90	V V	$R_L = 1MΩ$
Bandwidth	B_W		200			200			200		KHz	
Slew Rate	S_R		0.1			0.1			0.1		V/µs	$A_V = +1$ $C_L = 25pF$

$V_S = ±2.5V$ $−55°C ≤ T_A ≤ +125°C$ unless otherwise specified

| Parameter	Symbol	4706ADB			4706BDB			4706DB			Unit	Test Conditions
		Min	Typ	Max	Min	Typ	Max	Min	Typ	Max		
Input Offset Voltage	V_{OS}			3.0			6.0			15.0	mV	$R_S ≤ 1MΩ$
Input Offset Current	I_{OS}		1	4		1	4		1	4	nA	
Input Bias Current	I_B		1	4		1	4		1	4	nA	
Power Supply Rejection Ratio	PSRR	60	75		60	75		60	75		dB	$R_S ≤ 1MΩ$
Common Mode Rejection Ratio	CMRR	60	83		60	83		60	83		dB	$R_S ≤ 1MΩ$
Large Signal Voltage Gain	A_V	10	50		10	50		7	50		V/mV	$R_L = 1MΩ$
Output Voltage Range	V_O low V_O high	2.25	-2.40 2.40	-2.25	2.25	-2.40 2.40	-2.25	2.25	-2.40 2.40	-2.25	V V	$R_L = 1MΩ$

Design & Operating Notes:

1. The ALD4706A/ALD4706B/ALD4706 CMOS operational amplifier uses a 3 gain stage architecture and an improved frequency compensation scheme to achieve large voltage gain, high output driving capability, and better frequency stability. In a conventional CMOS operational amplifier design, compensation is achieved with a pole splitting capacitor together with a nulling resistor. This method is, however, very bias dependent and thus cannot accommodate the large range of supply voltage operation as is required from a stand alone CMOS operational amplifier. The ALD4706A/ALD4706B/ALD4706 is internally compensated for unity gain stability using a novel scheme that does not use a nulling resistor. This scheme produces a clean single pole roll off in the gain characteristics while providing for more than 70 degrees of phase margin at the unity gain frequency.

2. The ALD4706A/ALD4706B/ALD4706 has complementary p-channel and n-channel input differential stages connected in parallel to accomplish rail to rail input common mode voltage range. This means that with the ranges of common mode input voltage close to the power supplies, one of the two differential stages is switched off internally. To maintain compatibility with other operational amplifiers, this switching point has been selected to be about 1.5V below the positive supply voltage. Since offset voltage trimming on the ALD4706A/ALD4706B/ALD4706 is made when the input voltage is symmetrical to the supply voltages, this internal switching does not affect a large variety of applications such as an inverting amplifier or non-inverting amplifier with a gain larger than 2.5 (5V operation), where the common mode voltage does not make excursions above this switching point. The user should however, be aware that this switching does take place if the operational amplifier is connected as a unity gain buffer and should make provision in his design to allow for input offset voltage variations.

3. The input bias and offset currents are essentially input protection diode reverse bias leakage currents, and are typically less than 0.1pA at room temperature. This low input bias current assures that the analog signal from the source will not be distorted by input bias currents. Normally, this extremely high input impedance of greater than $10^{13}\Omega$ would not be a problem as the source impedance would limit the node impedance. However, for applications where source impedance is very high, it may be necessary to limit noise and hum pickup through proper shielding.

4. The output stage consists of class AB complementary output drivers, capable of driving a low resistance load. The output voltage swing is limited by the drain to source on-resistance of the output transistors as determined by the bias circuitry, and the value of the load resistor. When connected in the voltage follower configuration, the oscillation resistant feature, combined with the rail to rail input and output feature, makes an effective analog signal buffer for medium to high source impedance sensors, transducers, and other circuit networks.

5. The ALD4706A/ALD4706B/ALD4706 operational amplifier has been designed to provide full static discharge protection. Internally, the design has been carefully implemented to minimize latch up. However, care must be exercised when handling the device to avoid strong static fields that may degrade a diode junction, causing increased input leakage currents. In using the operational amplifier, the user is advised to power up the circuit before, or simultaneously with, any input voltages applied and to limit input voltages to not exceed 0.3V of the power supply voltage levels.

6. The ALD4706A/ALD4706B/ALD4706, with its ultra micropower operation, offers numerous benefits in reduced power supply requirements, less noise coupling and current spikes, less thermally induced drift, better overall reliability due to lower self heating, and lower input bias current. It requires practically no warm up time as the chip junction heats up to only 0.1°C above ambient temperature under most operating conditions.
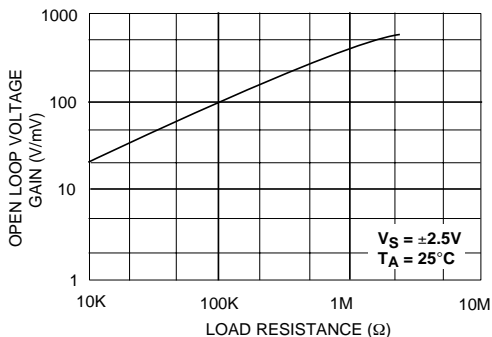
TYPICAL PERFORMANCE CHARACTERISTICS
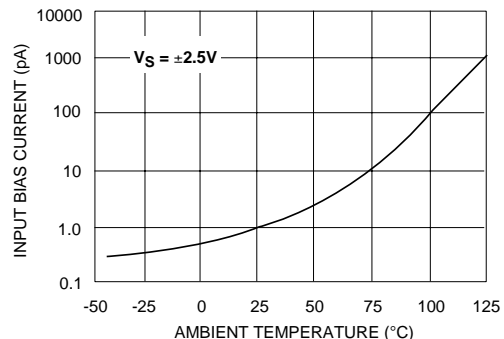
SUPPLY CURRENT AS A FUNCTION OF SUPPLY VOLTAGE
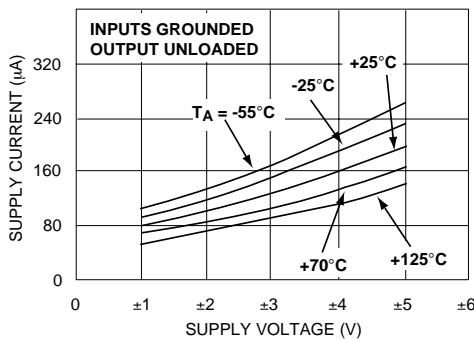
COMMON MODE INPUT VOLTAGE RANGE AS A FUNCTION OF SUPPLY VOLTAGE
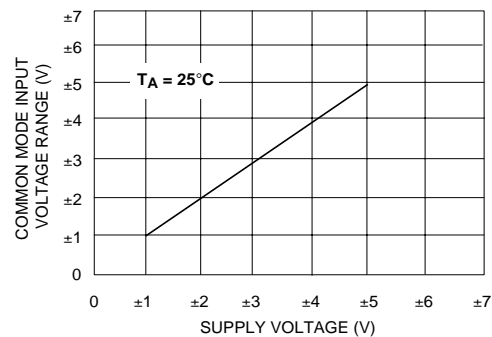
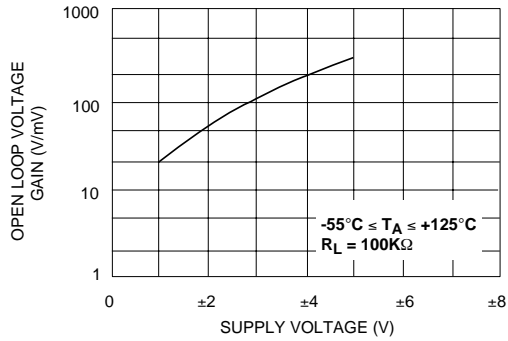### OPEN LOOP VOLTAGE GAIN AS A FUNCTION OF LOAD RESISTANCE

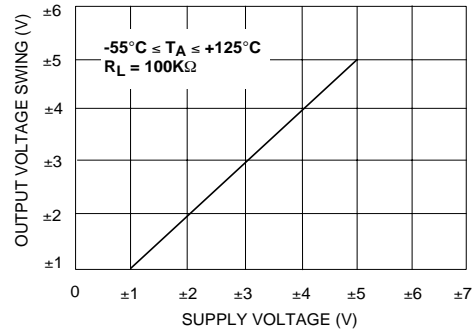### INPUT BIAS CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE

TYPICAL PERFORMANCE CHARACTERISTICS (cont'd)

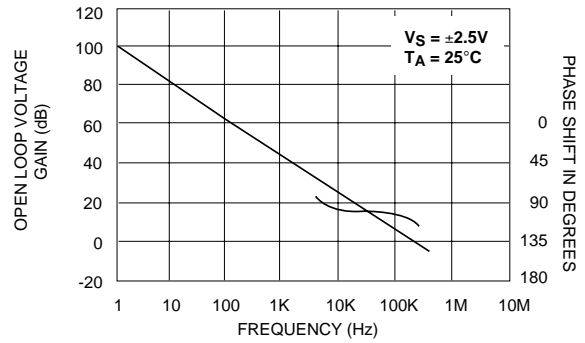### OPEN LOOP VOLTAGE GAIN AS A FUNCTION OF SUPPLY VOLTAGE AND TEMPERATURE

OPEN LOOP VOLTAGE GAIN (V/mV) vs SUPPLY VOLTAGE (V)

$-55°C \leq T_A \leq +125°C$
$R_L = 100K\Omega$

OUTPUT VOLTAGE SWING AS A FUNCTION OF SUPPLY VOLTAGE

OUTPUT VOLTAGE SWING (V) vs SUPPLY VOLTAGE (V)

$-55°C \leq T_A \leq +125°C$
$R_L = 100K\Omega$

INPUT OFFSET VOLTAGE AS A FUNCTION OF AMBIENT TEMPERATURE REPRESENTATIVE UNITS

INPUT OFFSET VOLTAGE (mV) vs AMBIENT TEMPERATURE (°C)

$V_S = ±2.5V$
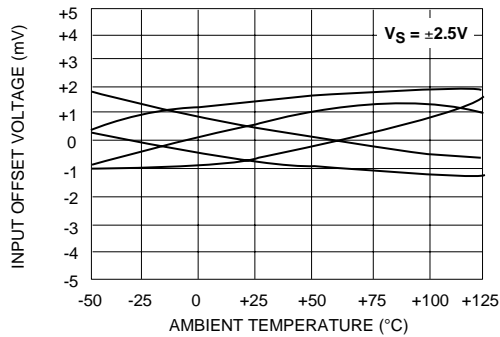
OPEN LOOP VOLTAGE GAIN AS A FUNCTION OF FREQUENCY

OPEN LOOP VOLTAGE GAIN (dB) / PHASE SHIFT IN DEGREES vs FREQUENCY (Hz)

$V_S = ±2.5V$
$T_A = 25°C$

INPUT OFFSET VOLTAGE AS A FUNCTION OF COMMON MODE INPUT VOLTAGE
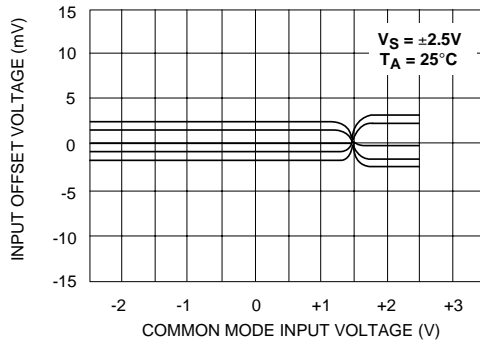
INPUT OFFSET VOLTAGE (mV) vs COMMON MODE INPUT VOLTAGE (V)

$V_S = ±2.5V$
$T_A = 25°C$

LARGE - SIGNAL TRANSIENT RESPONSE

2V/div

$V_S = ±1.0V$
$T_A = 25°C$
$R_L = 100K\Omega$
$C_L = 25pF$

500mV/div   10μs/div

LARGE - SIGNAL TRANSIENT RESPONSE

5V/div

$V_S = ±2.5V$
$T_A = 25°C$
$R_L = 100K\Omega$
$C_L = 25pF$

2V/div   10μs/div

SMALL - SIGNAL TRANSIENT RESPONSE

100mV/div

$V_S = ±2.5V$
$T_A = 25°C$
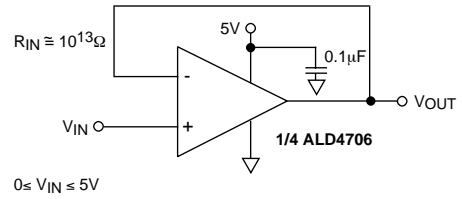$R_L = 100K\Omega$
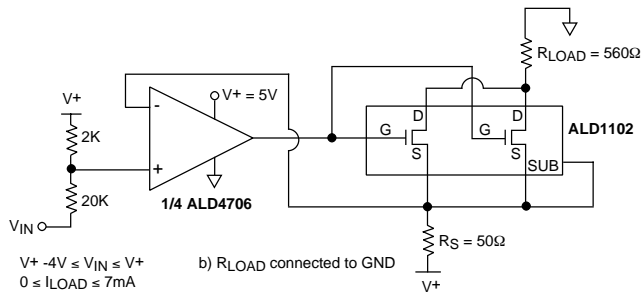$C_L = 25pF$

50mV/div   10μs/div
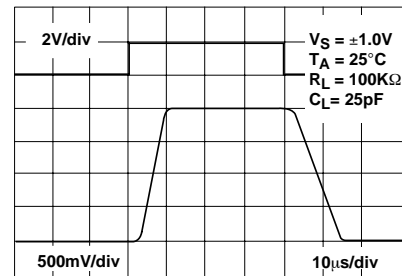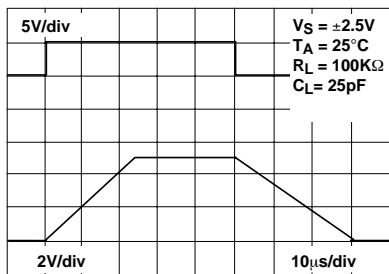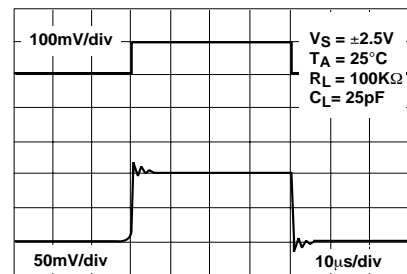
TYPICAL APPLICATIONS

V TO I AMPLIFIER
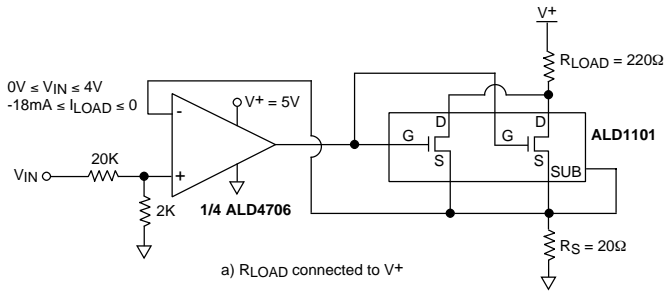
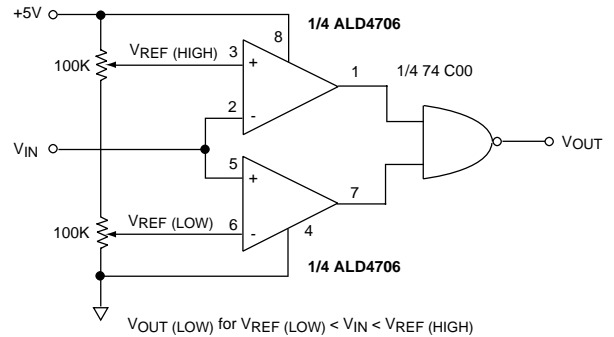### RAIL-TO-RAIL VOLTAGE FOLLOWER/BUFFER

V+

$R_{LOAD} = 220\Omega$

$0V \leq V_{IN} \leq 4V$
$-18mA \leq I_{LOAD} \leq 0$

V+ = 5V

ALD1101

V_{IN}

20K

2K

1/4 ALD4706

G D   G D
S   S   SUB

$R_S = 20\Omega$

a) R_{LOAD} connected to V+

$R_{IN} \cong 10^{13}\Omega$

5V

0.1µF

V_{IN}

1/4 ALD4706

V_{OUT}

$0 \leq V_{IN} \leq 5V$

V+ = 5V

$R_{LOAD} = 560\Omega$

V+

2K

20K

V_{IN}

1/4 ALD4706

ALD1102

G D   G D
S   S   SUB

$R_S = 50\Omega$

V+

V+ -4V $\leq V_{IN} \leq$ V+
$0 \leq I_{LOAD} \leq 7mA$

b) R_{LOAD} connected to GND

PHOTO DETECTOR CURRENT TO VOLTAGE CONVERTER

$R_F = 5M$

I

+2.5V

PHOTODIODE

$V_{OUT} = 1 \times R_F$

-2.5V

$R_L = 10K$

1/4 ALD4706
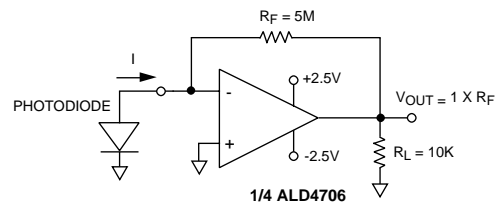
TWO STAGE HIGH GAIN AMPLIFIER

20K

20K

$1V \leq$ V+ $\leq 12V$

800Ω

1K

R_L 20K

C_L 1nF

V+

220Ω

V_{IN}

1.4µF

225Ω

1/4 ALD 4706

1/4 ALD4706

NOTES: 1) GAIN = 20 First Stage
2) GAIN = 25 Second Stage
3) 500Hz $\leq$ Operating Frequency $\leq$ 10KHz
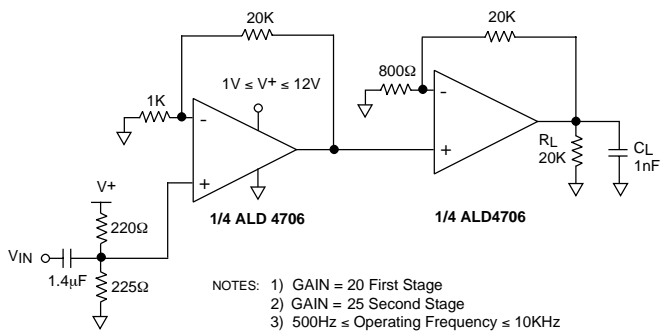
RAIL-TO-RAIL WINDOW COMPARATOR

+5V

1/4 ALD4706

100K

$V_{REF (HIGH)}$ 3

8

1

1/4 74 C00

2

V_{IN}

V_{OUT}

5

7

100K

$V_{REF (LOW)}$ 6

4

1/4 ALD4706

$V_{OUT (LOW)}$ for $V_{REF (LOW)} < V_{IN} < V_{REF (HIGH)}$

HIGH INPUT IMPEDANCE RAIL-TO-RAIL PRECISION DC SUMMING AMPLIFIER

V+ = +2.5V

0.1µF

V_1 10M

V_2 10M

10M

1/4 ALD4706

V_{OUT}

V_3 10M

0.1µF

V- $\leq V_{IN} \leq$ V+

V_4 10M

V- = - 2.5V

V- $\leq V_{OUT} \leq$ V+

10M

$V_{OUT} = V_1 + V_2 - V_3 - V_4$

$R_{IN} = 10M\Omega$ Accuracy limited by resistor tolerances and input offset voltage
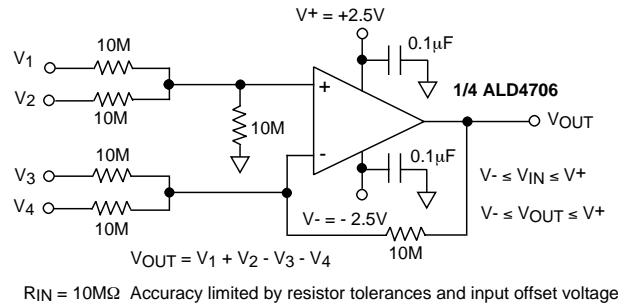
SOIC-14 PACKAGE DRAWING

14 Pin Plastic SOIC Package
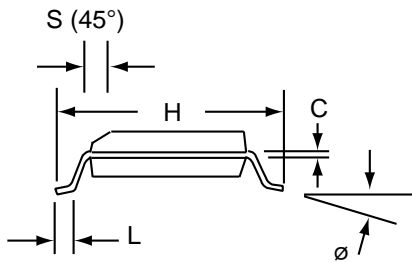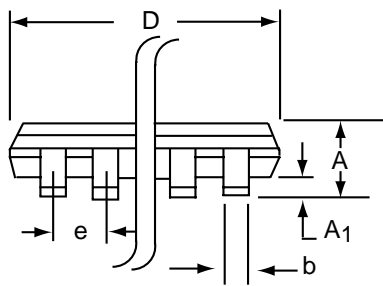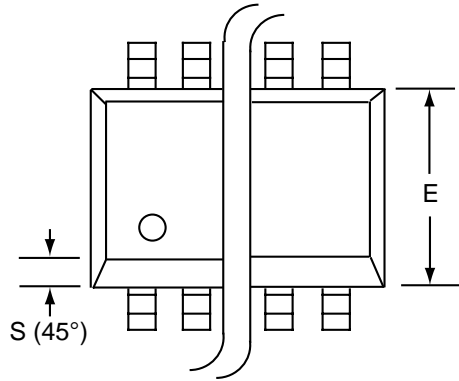
Dim	Millimeters		Inches	
	Min	**Max**	**Min**	**Max**
A	1.35	1.75	0.053	0.069
A$_1$	0.10	0.25	0.004	0.010
b	0.35	0.45	0.014	0.018
C	0.18	0.25	0.007	0.010
D-14	8.55	8.75	0.336	0.345
E	3.50	4.05	0.140	0.160
e	1.27 BSC		0.050 BSC	
H	5.70	6.30	0.224	0.248
L	0.60	0.937	0.024	0.037
Ø	0°	8°	0°	8°
S	0.25	0.50	0.010	0.020

PDIP-14 PACKAGE DRAWING

14 Pin Plastic DIP Package
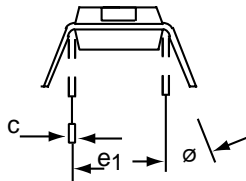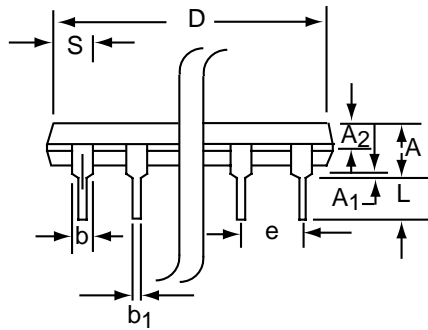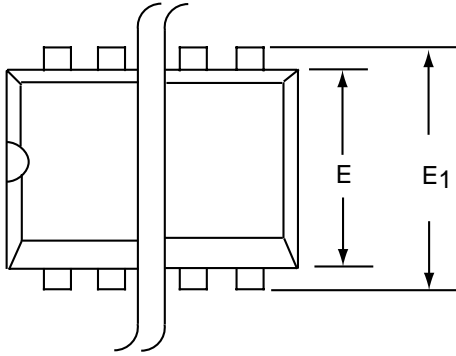
Dim	Millimeters		Inches	
	Min	Max	Min	Max
A	3.81	5.08	0.105	0.200
A_1	0.38	1.27	0.015	0.050
A_2	1.27	2.03	0.050	0.080
b	0.89	1.65	0.035	0.065
b_1	0.38	0.51	0.015	0.020
c	0.20	0.30	0.008	0.012
D-14	17.27	19.30	0.680	0.760
E	5.59	7.11	0.220	0.280
E_1	7.62	8.26	0.300	0.325
e	2.29	2.79	0.090	0.110
e_1	7.37	7.87	0.290	0.310
L	2.79	3.81	0.110	0.150
S-14	1.02	2.03	0.040	0.080
ø	0°	15°	0°	15°

CERDIP-14 PACKAGE DRAWING

14 Pin CERDIP Package
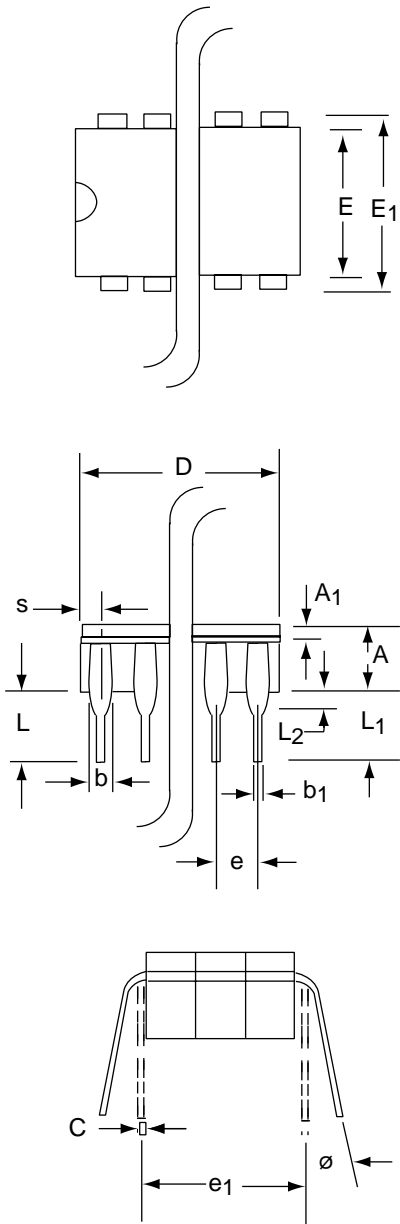
Dim	Millimeters		Inches	
	Min	**Max**	**Min**	**Max**
A	3.55	5.08	0.140	0.200
A$_1$	1.27	2.16	0.050	0.085
b	0.97	1.65	0.038	0.065
b$_1$	0.36	0.58	0.014	0.023
C	0.20	0.38	0.008	0.015
D-14	--	19.94	--	0.785
E	5.59	7.87	0.220	0.310
E$_1$	7.73	8.26	0.290	0.325
e	2.54 BSC		0.100 BSC	
e$_1$	7.62 BSC		0.300 BSC	
L	3.81	5.08	0.150	0.200
L$_1$	3.18	--	0.125	--
L$_2$	0.38	1.78	0.015	0.070
S	--	2.49	--	0.098
Ø	0°	15°	0°	15°

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