### Datasheet

### AS1361/AS1362 150mA/300mA, Ultra-Low-Noise, High-PSRR Low Dropout Regulators, with POK

### **1** General Description

The AS1361/AS1362 are ultra-low-noise, low-dropout linear regulators specifically designed to deliver up to 150/300mA continuous output current, and can achieve a low 140mV dropout for 300mA load current. The LDOs are designed and optimized to work with low-cost, small-capacitance ceramic capacitors.

The devices are available as the standard products listed in Table 1.

Table 1	. Standard	Products
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Model	Load Current	Output Voltage
AS1361	150mA	Preset – 1.5 to 4.5V
AS1362	300mA	Preset – 1.5 to 4.5V

An integrated P-channel MOSFET pass transistor allows the devices to maintain extremely low quiescent current  $(40\mu A)$ .

The AS1361/AS1362 uses an advanced architecture to achieve ultra-low output voltage noise of  $9\mu$ VRMS and a power-supply rejection-ratio of better than 80dB (up to 10kHz).

An active-Low, open-drain power-ok output indicates if the output voltage is within regulation.

The AS1361/AS1362 requires only  $1\mu$ F output capacitor for stability at any load. When the LDO is disabled, current consumption drops below 500nA.

The devices are available in a TSOT23 6-pin package.

### 2 Key Features

- Preset Output Voltages: 1.5 to 4.5V (in 50mV steps)
- Output Noise: 9µVRMs @ 100Hz to 100kHz
- Power-Supply Rejection Ratio: 92dB @ 1kHz
- Low Dropout: 140mV @ 300mA Load
- Power-OK Output
- Stable with 1µF Ceramic Capacitor for any Load
- Guaranteed 150/300mA Output
- 1.25V Internal Reference
- Extremely-Low Quiescent Current: 40µA
- Excellent Load/Line Transient
- Overcurrent and Thermal Protection
- TSOT23 6-pin Package

### **3** Applications

The devices are ideal for mobile phones, wireless phones, PDAs, handheld computers, mobile phone base stations, Bluetooth portable radios and accessories, wireless LANs, digital cameras, personal audio devices, and any other portable, battery-powered application.

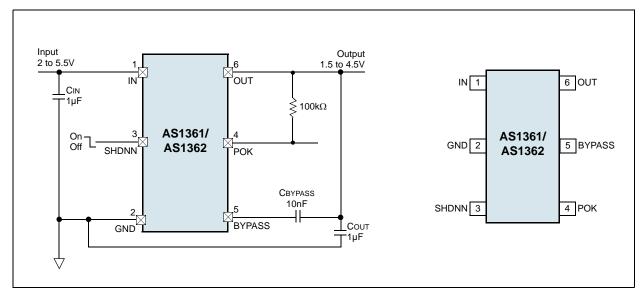
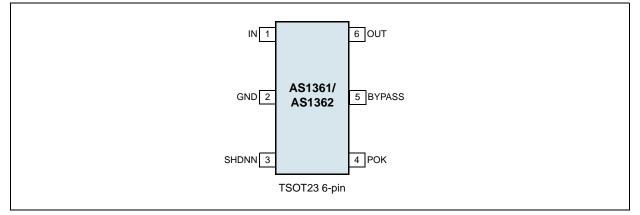


Figure 1. Typical Application Circuit

# 4 Pinout

### **Pin Assignments**

Figure 2. Pin Assignments (Top View)



### **Pin Descriptions**

Table 2. Pin Descriptions

Pin Number	Pin Name	Description
1	IN	Unregulated Input Supply.
2	GND	Ground
3	SHDNN	Shutdown. Pull this pin low to disable the LDO.
4	РОК	Power-OK Output. Active-Low, open-drain output indicates if the output voltage is within regulation. 0 = Vout <94% VoutNOM 1 = Vout >94% VoutNOM
5	BYPASS	Noise Bypass for Low-Noise Operation.Connect a 10nF capacitor fromthis pin to OUT.Note: This pin is shorted to GND in shutdown mode.
6	OUT	<b>Regulated Output Voltage.</b> Bypass this pin with a capacitor to GND. See Capacitor Selection and Regulator Stability on page 12 for more details.

# **5** Absolute Maximum Ratings

Stresses beyond those listed in Table 3 may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in Electrical Characteristics on page 4 is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Table 3.	Absolute	Maximum	Ratings
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Parameter	Min	Max	Units	Comments
IN to GND	-0.3	+7	V	
OUT, SHDNN, POK to GND	-0.3	IN + 0.3	V	
BYPASS to GND	-0.3	OUT + 0.3	V	
Output Short-Circuit Duration		Infinite		
Thermal Resistance <sub>OJA</sub>		201.7	°C/W	on PCB
Operating Temperature Range	-40	+85	°C	
Junction Temperature		+150	°C	
Storage Temperature Range	-65	+150	°C	
Package Body Temperature		+260	Ŷ	The reflow peak soldering temperature (body temperature) specified is in accordance with IPC/JEDEC J-STD-020D "Moisture/Reflow Sensitivity Classification for Non-Hermetic Solid State Surface Mount Devices". The lead finish for Pb-free leaded packages is matte tin (100% Sn).

## **6 Electrical Characteristics**

 $V_{IN} = V_{OUT} + 0.5V$ ,  $C_{IN} = 1\mu F$ ,  $C_{OUT} = 1\mu F$ ,  $C_{BYPASS} = 10nF$ ,  $T_{AMB} = -40$  to  $+85^{\circ}C$  (unless otherwise specified). Typ values are at  $T_{AMB} = +25^{\circ}C$ . Limits 100% tested at  $+25^{\circ}C$ . Limits over operating temperature range guaranteed by design. Table 4. Electrical Characteristics

Symbol	Parameter	Condition	Min	Тур	Max	Unit			
Vin	Input Voltage Range		2		5.5	V			
		IOUT = 1mA, TAMB = +25°C	-0.5		+0.5				
		IOUT = 100µA to 150mA, Тамв = +25°C (AS1361)	-0.75		+0.75				
	Output Voltage Accuracy	IOUT = 100µA to 300mA, Тамв = +25°C (AS1362)	-1.0		+1.0	+1.0 %			
		IOUT = 100μA to 150mA, (AS1361)	-1.5		+1.5	-			
		IOUT = 100µA to 300mA, (AS1362)	-2.0		+2.0				
lour.	Maximum Quitaut Quimant	AS1361	150						
Ιουτ	Maximum Output Current	AS1362	300			mA			
h n a <del>r</del>	Current Limit	AS1361, OUT = 90% of nom., TAMB = +25°C		270					
ILIMIT	Current Limit	AS1362, OUT = 90% of nom., TAMB = +25°C		510		mA			
		Vout $\ge$ 3V, lout = 150mA		70	95				
		Vout $\geq$ 3V, lout = 300mA, AS1362		140	200				
	1	$2.5V \le VOUT < 3V$ , IOUT = 150mA		90	120	mV			
Dropout Vo	Dropout Voltage <sup>1</sup>	$2.5V \le VOUT < 3V$ , IOUT = 300mA, AS1362		170	230				
		$2.0V \le VOUT < 2.5V$ , IOUT = 150mA		140	190				
		$2.0V \leq V \text{OUT} < 2.5V \text{, IOUT} = 300 \text{mA} \text{, AS1362}$		270	350				
lo	Quiecoant Current	IOUT = 0.05mA		40	90				
lq	Quiescent Current	VIN = VOUTNOM - 0.1V, IOUT = 0mA		150	250	μA			
Vlnr	Line Regulation	VIN = (VOUT +0.5V) to 5.5V, IOUT = 0.1mA		0.02		%/V			
Vldr	Load Regulation	IOUT = 1 to 150/300mA		0.0005		%/mA			
ISHDNN	Shutdown Supply Current	SHDNN = 0V		9	500	nA			
		f = 1kHz, IOUT = 10mA		92					
PSRR	Ripple Rejection	f = 10kHz, Iout = 10mA		80		dB			
		f = 100kHz, IOUT = 10mA		62					
	Output Noise Voltage (RMS)	f = 100Hz to 100kHz, ILOAD = 0 to 150/300mA		9		μV			
Shutdov	'n								
	Shutdown Exit Delay <sup>2</sup>	$R$ LOAD = 50 $\Omega$			300	μs			
	SHDNN Logic Low Level	VIN = 2 to 5.5V			0.4	V			
	SHDNN Logic High Level	VIN = 2 to 5.5V	1.5			V			
Power-C	K Output								
Vрок	Power-OK Voltage Threshold	IOUT = 0, VOUTRISING	90	94	98	%			
VFUN	i ower-on vollage mieshold	Hysteresis, IOUT = 0		1.5		Vout			
Vol	POK Output Voltage Low	ISINK = 1mA			0.3	V			
	POK Output Leakage Current	Vou⊤ in regulation			1	μA			

Table 4.	Electrical	Characteristics	(Continued)
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Symbol	Parameter	Condition	Min	Тур	Max	Unit
Thermal	Protection					
TSHDNM	Thermal Shutdown Temperature			160		٥C
∆Tshdn M	Thermal Shutdown Hysteresis			15		°C

1. Dropout is defined as VIN - VOUT when VOUT is 100mV below the value of VOUT for VIN = VOUT + 0.5V.

2. Time needed for VOUT to reach 90% of final value.

**Note:** All limits are guaranteed. The parameters with min and max values are guaranteed with production tests or SQC (Statistical Quality Control) methods.



## **7** Typical Operating Characteristics

VIN = VOUT + 0.5V, CIN = COUT = 1µF, CBYPASS = 10nF, TAMB = 25°C (unless otherwise specified).

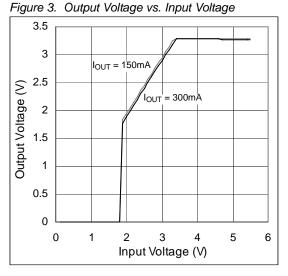


Figure 5. Output Voltage Accuracy vs. Temperature

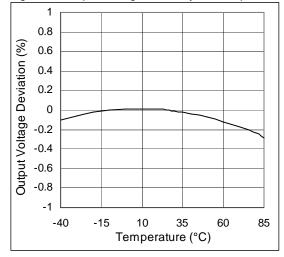


Figure 7. Dropout Voltage vs. Output Voltage

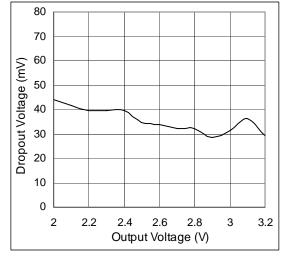


Figure 4. Output Voltage Accuracy vs. Load Current

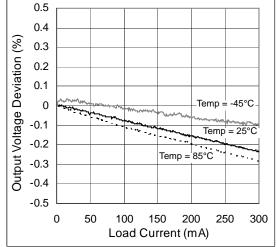
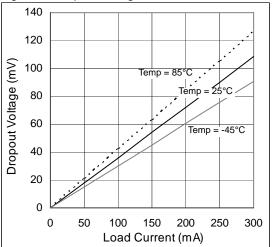


Figure 6. Dropout Voltage vs. Load Current



#### Figure 8. Ground Pin Current vs. Input Voltage

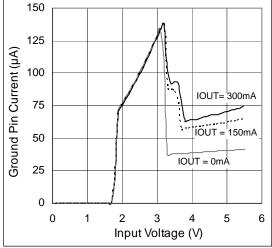
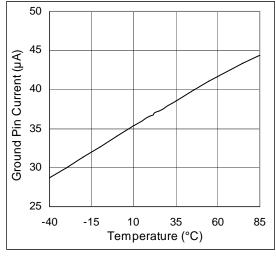
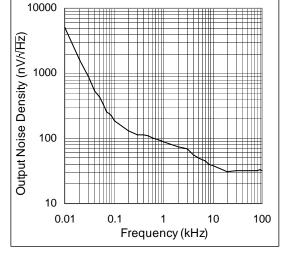


Figure 10. Ground Pin Current vs. Temperature







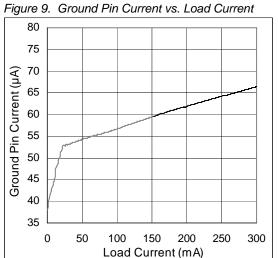


Figure 11. PSRR vs. Frequency; IOUT = 10mA

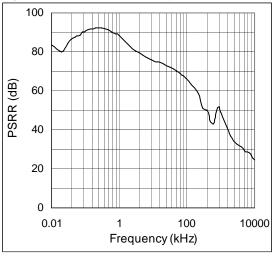


Figure 13. Output Noise vs. Bypass Capacitance

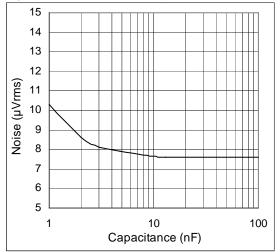


Figure 14. Load Transient Response,  $V_{IN} = 3.8V$ ,  $V_{OUT} = 3.3V$ 

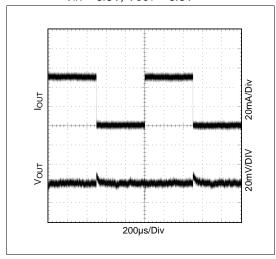


Figure 16. Line Transient Response

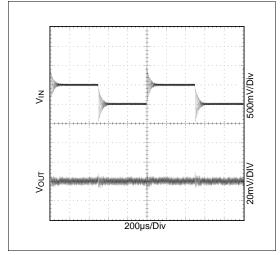


Figure 15. Load Transient Response near Dropout,  $V_{IN} = 3.4V$ ,  $V_{OUT} = 3.3V$ 

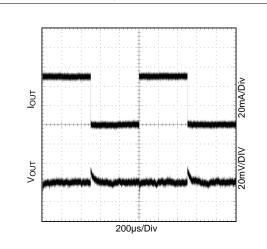
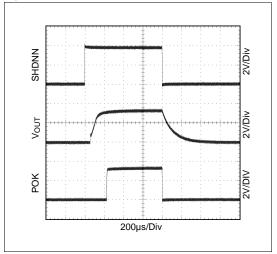


Figure 17. Enter & Exit Shutdown Delay



## 8 Detailed Description

The AS1361/AS1362 are ultra-low-noise, low-dropout, low-quiescent current linear-regulators specifically designed for space-limited applications. The devices are available with preset output voltages from 1.5 to 4.5V in 50mV increments.

These devices can supply loads up to 150/300mA. As shown in Figure 18, the AS1361/AS1362 consist of an integrated bandgap core and noise bypass circuitry, error amplifier, P-channel MOSFET pass transistor, and internal feedback voltage-divider.

The output voltage is fed back through an internal resistor voltage-divider connected to pin OUT. An external bypass capacitor connected to pin BYPASS reduces noise at the output. Additional blocks include a current limiter, thermal sensor, and shutdown logic.

#### **Internal Voltage Reference**

The 1.25V bandgap reference is connected to the error amplifier's inverting input. The error amplifier compares this reference with the feedback voltage and amplifies the difference. If the feedback voltage is lower than the reference voltage, the pass-transistor gate is pulled low. This allows more current to pass to the output and increases the output voltage. If the feedback voltage is too high, the pass transistor gate is pulled high, allowing less current to pass to the output.

#### **Internal P-Channel Pass Transistor**

The AS1361/AS1362 feature a  $0.5\Omega$  (typ) P-channel MOSFET pass transistor, which provides several advantages over similar designs using a PNP pass transistor, including prolonged battery life. The P-channel MOSFET does not require a base driver, thus quiescent current is dramatically reduced. The AS1361/AS1362 LDOs do not exhibit problems associated with typical PNP-based LDOs, and consume only 40µA of quiescent current in light load and 220µA in dropout (see Typical Operating Characteristics on page 6).

#### **Output Voltage**

The AS1361/AS1362 deliver preset output voltages from 1.5 to 4.5V, in 50mV increments (see Ordering Information on page 14).

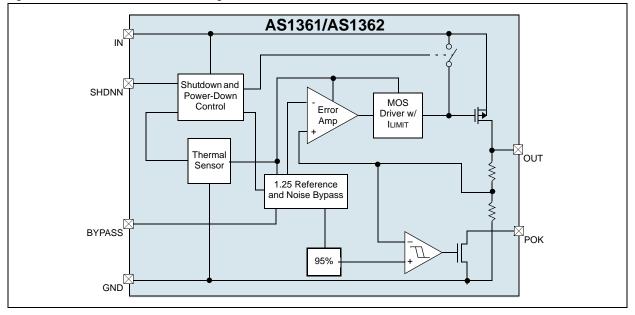
#### Shutdown

The AS1361/AS1362 feature a low-power shutdown mode that reduces quiescent current to <200nA. Driving SHDNN low disables the internal voltage reference, error amplifier, gate-drive circuitry, and P-channel MOSFET pass transistor (see Figure 18), and the device output enters a high-impedance state.

Note: For normal operation connect pin SHDNN to pin IN.

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Figure 18. AS1361/AS1362 - Block Diagram
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#### Power-OK

The power-ok circuitry is built around an N-channel MOSFET. The circuitry monitors the output voltage and if the voltage goes out of regulation (e.g. during dropout, current limit, or thermal shutdown) pin POK goes low.

The power-OK feature is not active during shutdown and provides a power-on-reset (POR) function that can operate down to  $V_{IN} = 1V$ . A capacitor to GND may be added to generate a POR delay.

To obtain a logic-level output, connect a pull-up resistor from pin POK to pin OUT. Larger values for this resistor will help minimize current consumption; a  $100k\Omega$  resistor is perfect for most applications (see Figure 1 on page 1).

#### **Current Limit**

The AS1361/AS1362 include a current limiting circuitry to monitor and control the P-channel MOSFET pass transistor's gate voltage, thus limiting the device output current to 270mA (AS1361) and 510mA (AS1362).

**Note:** See Table 4 on page 4 for the recommended min and max current limits. The output can be shorted to ground indefinitely without causing damage to the device.

#### **Thermal Protection**

Integrated thermal protection circuitry limits total power dissipation in the AS1361/AS1362. When the junction temperature (TJ) exceeds +160°C, the thermal sensor signals the shutdown logic, turning off the P-channel MOSFET pass transistor and allowing the device to cool down. The thermal sensor turns the pass transistor on again after the device's junction temperature drops by 10°C, resulting in a pulsed output during continuous thermal-overload conditions.

**Note:** Thermal protection is designed to protect the devices in the event of fault conditions. For continuous operation, do not exceed the absolute maximum junction temperature rating of +150°C.

#### **Operating Region and Power Dissipation**

The AS1361/AS1362 maximum power dissipation is dependent on the thermal resistance of the case and PCB, the temperature difference between the die junction and TAMB, and airflow rate. Power dissipation across the device is calculated as:

$$PD = IOUT (VIN - VOUT)$$
 (EQ 1)

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The maximum power dissipation is calculated:

$$PDMAX = (T_J - T_{AMB})/(\theta_{JC} + \theta_{CA})$$
(EQ 2)

#### Where:

TJ - TAMB is the temperature difference between the AS1361/AS1362 die junction and the surrounding air;

 $\theta_{JC}$  is the thermal resistance of the package;

 $\theta$ CA is the thermal resistance through the PC board/copper traces/other materials to the surrounding air.

**Note:** Pin GND of the AS1361/AS1362 provides the electrical connection to system ground and also serves as a heat sink. Connect pin GND to the system ground using a large pad or ground plane.

#### **Noise Reduction**

The AS1361/AS1362 noise bypass circuitry dramatically reduces output noise, exhibiting  $9\mu$ VRMS of output voltage noise with CBYPASS =  $0.01\mu$ F and COUT =  $1\mu$ F. Use an external  $0.01\mu$ F bypass capacitor between pin BYPASS and pin OUT (see Figure 1 on page 1).

Note: Startup time is minimized by internal power-on circuitry which pre-charges CBYPASS.

## **9** Application Information

#### **Capacitor Selection and Regulator Stability**

For normal operation, use a 1µF capacitor at pin IN and a 1µF capacitor at pin OUT. Larger input capacitor values and lower ESR provide better noise rejection and line-transient response. Reduce output noise and improve load-transient response, stability, and power-supply rejection by using large output capacitors.

**Note:** Some ceramic dielectrics exhibit large capacitance and ESR variation with temperature. With dielectrics such as Z5U and Y5V, it may be necessary to use a 2.2μF or larger output capacitor to ensure stability at temperatures below -10°C. With X7R or X5R dielectrics, 1μF is sufficient at all operating temperatures.

#### **Bypass Capacitor**

Use a  $0.01\mu$ F bypass capacitor at pin BYPASS for low-output voltage noise reduction. The leakage current going into pin BYPASS should be less than 10nA. Increasing the capacitance slightly decreases the output noise. Values above  $0.1\mu$ F and below  $0.001\mu$ F are not recommended.

#### Noise, PSRR, and Transient Response

The AS1361/AS1362 are designed to deliver ultra-low noise and high PSRR, as well as low dropout and low quiescent currents in battery-powered systems. The power-supply rejection is 92dB at 1kHz and 62dB at 100kHz. (see PSRR vs. Frequency; IOUT = 10mA on page 7).

When operating from sources other than batteries, improved supply-noise rejection and transient response can be achieved by increasing the values of the input and output capacitors, and through passive filtering techniques.

The Figure 16 and Figure 14 on page 8 show the AS1361/AS1362 line- and load-transient responses.

#### **Dropout Voltage**

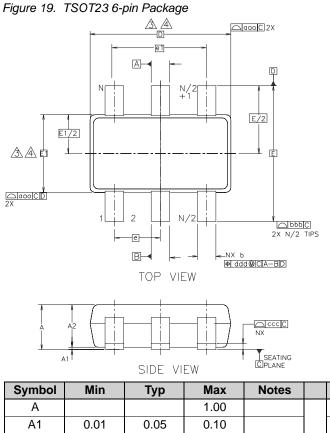
The AS1361/AS1362 minimum dropout voltage determines the lowest usable supply voltage. In battery-powered systems, this determines the useful end-of-life battery voltage.

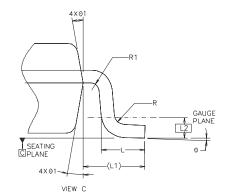
Since the AS1361/AS1362 use a P-channel MOSFET pass transistor, the dropout voltage is a function of drain-tosource on-resistance (RDS(ON)) multiplied by ILOAD (see Figure 6 on page 6).

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# **10 Package Drawings and Markings**

The devices are available in a TSOT23 6-pin package.





Symbol	Min	Тур	Max	Notes		Symbol	Min	Тур	Max	Notes
Α			1.00			L	0.30	0.40	0.50	
A1	0.01	0.05	0.10			L1		0.60REF		
A2	0.84	0.87	0.90			L2		0.25BSC		
b	0.30		0.45			N		6		
b1	0.31	0.35	0.39			R	0.10			
С	0.12	0.15	0.20			R1	0.10		0.25	
c1	0.08	0.13	0.16			θ	0°	4º	8º	
D		2.90BSC		3,4		θ1	4º	10º	12º	
E		2.80BSC		3,4			Tolerances	of Form a	nd Position	1
E1		1.60BSC		3,4		aaa		0.15		
е		0.95BSC			1	bbb		0.25		
e1		1.90BSC			1	CCC		0.10		
					1	ddd		0.20		

#### Notes:

- 1. Dimensioning and tolerancing conform to ASME Y14.5M 1994.
- 2. Dimensions are in millimeters.
- 3. Dimension D does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, and gate burrs shall not exceed 0.15mm per end. Dimension E1 does not include interlead flash or protrusion. Interlead flash or protrusion shall not exceed 0.15mm per side. Dimensions D and E1 are determined at datum H.
- 4. The package top can be smaller than the package bottom. Dimensions D and E1 are determined at the outermost extremes of the plastic body exclusive of mold flash, tie bar burrs, gate burrs, and interlead flash, but include any mistmatches between the top of the package body and the bottom. D and E1 are determined at datum H.

# **11 Ordering Information**

The devices are available as the standard products shown in Table 5.

Table 5. Ordering Ini		Output Output	Outrout Maltana	Delleren Ferme	Dealers
Ordering Code	Marking	Output Current	Output Voltage	Delivery Form	Package
AS1361-BTTT-15	ASL2	150mA	1.5V	Tape and Reel	TSOT23 6-pin
AS1361-BTTT-18	ASL3	150mA	1.8V	Tape and Reel	TSOT23 6-pin
AS1361-BTTT-25	ASL4	150mA	2.5V	Tape and Reel	TSOT23 6-pin
AS1361-BTTT-26	ASL5	150mA	2.6V	Tape and Reel	TSOT23 6-pin
AS1361-BTTT-27	ASL6	150mA	2.7V	Tape and Reel	TSOT23 6-pin
AS1361-BTTT-28	ASL7	150mA	2.8V	Tape and Reel	TSOT23 6-pin
AS1361-BTTT-285	ASL8	150mA	2.85V	Tape and Reel	TSOT23 6-pin
AS1361-BTTT-30	ASL9	150mA	3.0V	Tape and Reel	TSOT23 6-pin
AS1361-BTTT-33	ASMA	150mA	3.3V	Tape and Reel	TSOT23 6-pin
AS1361-BTTT-45	ASMB	150mA	4.5V	Tape and Reel	TSOT23 6-pin
AS1362-BTTT-15	ASMC	300mA	1.5V	Tape and Reel	TSOT23 6-pin
AS1362-BTTT-18	ASMD	300mA	1.8V	Tape and Reel	TSOT23 6-pin
AS1362-BTTT-25	ASME	300mA	2.5V	Tape and Reel	TSOT23 6-pin
AS1362-BTTT-26	ASMF	300mA	2.6V	Tape and Reel	TSOT23 6-pin
AS1362-BTTT-27	ASMG	300mA	2.7V	Tape and Reel	TSOT23 6-pin
AS1362-BTTT-28	ASMH	300mA	2.8V	Tape and Reel	TSOT23 6-pin
AS1362-BTTT-285	ASMI	300mA	2.85V	Tape and Reel	TSOT23 6-pin
AS1362-BTTT-30	ASMJ	300mA	3.0V	Tape and Reel	TSOT23 6-pin
AS1362-BTTT-33	ASMK	300mA	3.3V	Tape and Reel	TSOT23 6-pin
AS1362-BTTT-45	ASML	300mA	4.5V	Tape and Reel	TSOT23 6-pin

Table 5. Ordering Information

Non-standard devices from 1.5V to 4.5V are available in 50mV steps. For more information and inquiries contact http://www.austriamicrosystems.com/contact

**Note:** All products are RoHS compliant and Pb-free.

Buy our products or get free samples online at ICdirect: http://www.austriamicrosystems.com/ICdirect

For further information and requests, please contact us mailto:sales@austriamicrosystems.com or find your local distributor at http://www.austriamicrosystems.com/distributor

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