

# **AS5013**

# **Low Power Integrated Hall IC for Human Interface Applications**

# **General Description**

The AS5013 is a complete Hall Sensor IC for smart navigation key applications to meet the low power requirements and host SW integration challenges for products such as cell phones and smart handheld devices.

Due to the on chip processing engine, system designers are not tasked with integrating complex SW algorithms on their host processor thus leading to rapid development cycles.

The AS5013 single-chip IC includes 5 integrated Hall sensing elements for detecting up to  $\pm 2$ mm lateral displacement, high resolution ADC, XY coordinate and motion detection engine combined with a smart power management controller.

The X and Y positions coordinates and magnetic field information for each Hall sensor element is transmitted over a 2-wire I<sup>2</sup>C compliant interface to the host processor.

The AS5013 is available in a small 16-pin 4mmx4mmx0.55mm QFN package and specified over an operating temperature of  $-20^{\circ}\text{C}$  to  $80^{\circ}\text{C}$ 

Ordering Information and Content Guide appear at end of datasheet.

# **Key Benefits & Features**

The benefits and features of AS5013, Low Power Integrated Hall IC for Human Interface Applications are listed below:

Figure 1: Added Value of Using AS5013

Benefits	Features
Lateral magnet movement radius up to 2mm	8-bit resolution in X and Y direction
Fast and simple interface integration	High-speed I <sup>2</sup> C interface
Automatic low-power-mode	• 3μA current consumption in stand by
Low operating voltage supply	2.7V to 3.6V operating voltage - Down to 1.7V peripheral supply voltage
Contactless push-button possible	12-bit raw hall output
Small package, lower cost	QFN4x4 package
Push button and Motion detection	Various interrupt modes available
Insensitive to external magnetic stray fields	



# **Applications**

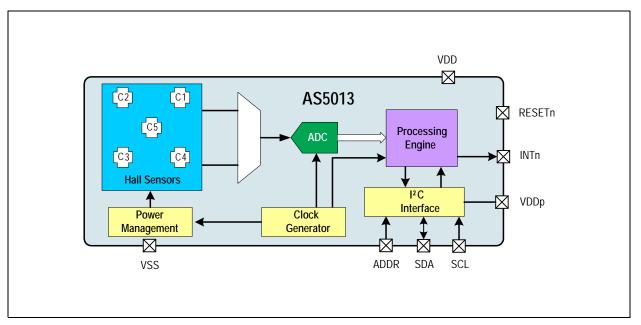
The AS5013 is ideal for small form-factor manual input devices in battery operated equipment, such as

- Mobile phones
- MP3 players
- PDAs
- GPS receivers
- Gaming consoles

# **Block Diagram**

The functional blocks of this device are shown below:

Figure 2: AS5013 Block Diagram



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# **Pin Assignments**

Figure 3: Pin Diagram (Top View)

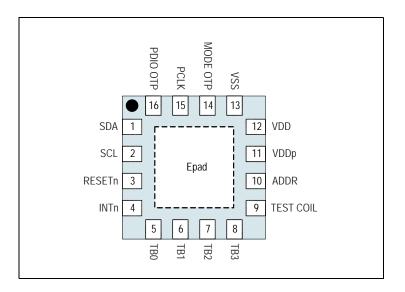


Figure 4: Pin Descriptions

Pin Number	Pin Name	Pin Type	ESD	Description
1	SDA	Digital I/O /Open drain	2kV	I <sup>2</sup> C data line, open drain
2	SCL		2kV	I <sup>2</sup> C clock line
3	RESETn	Digital input	2kV	General reset input 0: Reset 1: Normal mode
4	INTn	Digital output open drain	2kV	Interrupt line, open drain, active low
5	TB0		2kV	
6	TB1	Analog I/O	2kV	Test pin, leave unconnected
7	TB2	Analog I/O	2kV	rest pin, leave unconnected
8	TB3		2kV	
9	TEST COIL	Special	2kV Test pin, leave unconnected connect to VSS	
10	ADDR	Digital input with Schmitt trigger functionality	2kV	I <sup>2</sup> C address selection input. Read in at each reset

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Pin Number	Pin Name	Pin Type	ESD	Description
11	VDDp		2kV	1.7 ~ 3.6V IO power supply
12	VDD	Supply pad	2kV	2.7 ~ 3.6V core power supply
13	VSS		2kV	Power supply ground
14	MODE OTP		2kV	
15	PCLK	Digital I/O	2kV	Test pin, leave unconnected
16	PDIO OTP		2kV	
EPAD	Exposure Pad	-	-	Internally not connected. Leave open or connect to VSS

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# **Absolute Maximum Ratings**

Stresses beyond those listed in Figure 5 may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in Operating Conditions is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Figure 5: **Absolute Maximum Ratings** 

Symbol	Parameter	Min	Max	Units	Comments						
Electrical Parameters											
VDD	DC supply voltage	-0.3	5	V							
VDDp	Peripheral supply voltage	-0.3	5 VDD +0.3	V							
Vin	Input pin voltage	-0.3	VDDp +0.3	V							
VIII	input pin voitage	-	3.6	V							
l <sub>scr</sub>	Input current (latchup immunity)	±	100	mA	JEDEC 78						
		Electros	tatic Discharg	е							
ESD	Electrostatic discharge	Ξ	±2	kV	MIL 883 E method 3015, direct pad contact						
$\Theta_{JA}$	Package thermal resistance	-	32	K/W	Velocity=0, Multi Layer PCB; JEDEC Standard Testboard						
	Tempe	rature Rang	es and Storage	e Conditio	ns						
T <sub>strg</sub>	Storage temperature	-55	125	°C							
T <sub>body</sub>	Package body temperature		260	°C	IPC/JEDEC J-STD-020 (1)						
RH <sub>NC</sub>	Humidity non-condensing	5	85	%							
MSL	Moisture Sensitive Level		3		Represents a maximum floor life time of 168h						

### Note(s):

1. The reflow peak soldering temperature (body temperature) specified is in accordance with IPC/JEDEC J-STD-020 "Moisture/Reflow Sensitivity Classification for Non-Hermetic Solid State Surface Mount Devices". The lead finish for Pb-free leaded packages is matte tin (100% Sn).

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# **Electrical Characteristics**

# **Operating Conditions**

 $T_{AMB} = -20$ °C to 80°C, VDD = 3.3V, RESETn = HIGH

Figure 6: **Operating Conditions** 

Symbol	Parameter	Conditions	Min	Тур	Max	Units
VDD	Core supply voltage		2.7		3.6	V
VDDp	Peripheral supply voltage	Input: RESETn Open drain outputs: SCL, SDA, INTn. External I <sup>2</sup> C pull up resistor to be connected to VDDp.	1.7		VDD	V
	Maximal average current consumption on VDD,	$T_{AMB} = -20$ °C to 50°C	3+	3760/ts	[ms]	
IDD <sub>s</sub>	Pulsed peaks = IDD <sub>f</sub> depends on the sampling time ts[ms]	T <sub>AMB</sub> = 50°C to 80°C	10+	-3760/ts	[ms]	μΑ
IDD <sub>I</sub>	Current consumption on core supply, Idle mode,	T <sub>AMB</sub> = -20°C to 50°C			3	
וטטן	no readout (ts = infinite)	T <sub>AMB</sub> = 50°C to 80°C			10	μΑ
IDD <sub>f</sub>	Current consumption on core supply, Idle mode, continuous readout (ts=450µs)	Continuous current pin VDD Maximum sampling ts = 450µs			10	mA
Tpua	Power up time analog	Step on VDD to Data_Ready			1000	μs
Tconv	Conversion time	Read X/Y coordinate I <sup>2</sup> C Y_ res_int ACK bit of to Data_ Ready			450	μs
t <sub>P,W</sub>	Nominal wakeup time		20		320	ms
dx dy	Lateral movement radius	The range depends on the magnet and the distance to the surface, dx <sup>2</sup> +dy <sup>2</sup> <= 4mm			2	mm
d	Type of magnet	Cylindrical; axial magnetized	2		3	mm
RH	Hall array diameter			2.2		mm
B <sub>Z</sub>	Magnetic field strength	Vertical magnetic field at magnet center; measured at chip surface	30		120	mT
T <sub>AMB</sub>	Ambient temperature range		-20		80	°C

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Symbol	Parameter	Conditions	Min	Тур	Max	Units
	Resolution of XY displacement	Over 2*dx and 2*dy axis		8		bit
	Noise (RMS)	C1C5 channel data (result from two measurement – positive and negative current spinning)			100	μТ
PSSR	Power Supply Rejection Ratio	VDD=3.3V; Temp = 25°C dVDD= 100 mVpp at 10.30kHz			0.2	%/100 mV
	IC package		4mm>	QFN16 4mmx4mmx0.55mm		
	Power supply filtering	Ceramic capacitor VDD - VSS	100			nF
	capacitors	Ceramic capacitor VDDp - VSS	100			nF

# **Digital IO pads DC/AC Characteristics**

Figure 7: DC/AC Characteristics

Symbol	Parameter Conditions Min		Min	Max	Units					
	Inputs: SCL, SDA									
V <sub>IH</sub>	High level input voltage	IIC	0.7 * VDDp		V					
V <sub>IL</sub>	Low level input voltage	IIC		0.3 * VDDp	V					
I <sub>LEAK</sub>	Input leakage current	VDDp = 3.6V		1	μΑ					
	Inpu	uts: ADDR, RESETn (JED	EC76)							
V <sub>IH</sub>	High level input voltage	JEDEC	0.65 * VDDp		V					
V <sub>IL</sub>	Low level input voltage	JEDEC		0.35 * VDDp	V					
I <sub>LEAK</sub>	Input leakage current	VDDp = 3.6V		1	μΑ					

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Symbol	Parameter	Conditions	Min	Max	Units					
Outputs: SDA										
V <sub>OH</sub>	High level output voltage	High level output voltage	Open drain		Leakage current 1µA					
V <sub>OL1</sub>		-6mA; VDDP > 2V; Fast mode		VSS + 0.4	V					
V <sub>OL3</sub>	Low lovel output voltage	-6mA; VDDP ≤ 2V; Fast mode		VDDP*0.2	V					
V <sub>OL1</sub>		-3mA; VDDP > 2V; High speed		VSS + 0.4	V					
V <sub>OL3</sub>		-3mA; VDDP ≤ 2V; high speed		VDDP*0.2	V					
		Standard mode (100kHz)		400	pF					
C <sub>L</sub>	Capacitive load	Fast mode (400kHz)		400	pF					
		High speed mode (3.4MHz)		100	pF					
		Outputs: INTn (JEDEC7	6)							
V <sub>OH</sub>	High level output voltage	High level output voltage	Open drain		Leakage current 1µA					
V <sub>OL</sub>		-100μΑ		VSS + 0.2	V					
V <sub>OL</sub>	Low level output voltage	-2mA		VSS + 0.45	V					
C <sub>L</sub>	Capacitive load	Standard mode (100kHz)		30	pF					

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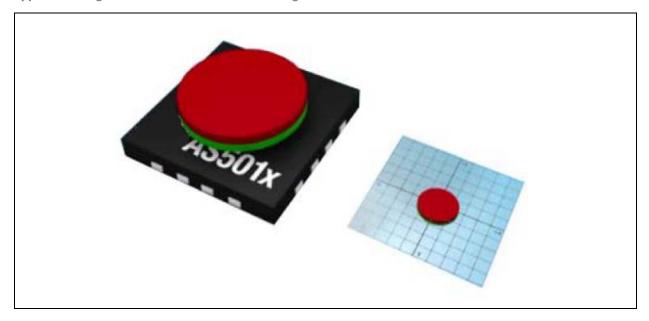


# **Detailed Description**

The benefits of the AS5013 device are as follows:

- Complete system-on-chip
- High reliability due to non-contact sensing
- Low power consumption

Figure 8:
Typical Arrangement of AS5013 and Axial Magnet



# Operating the AS5013

#### **Typical Application**

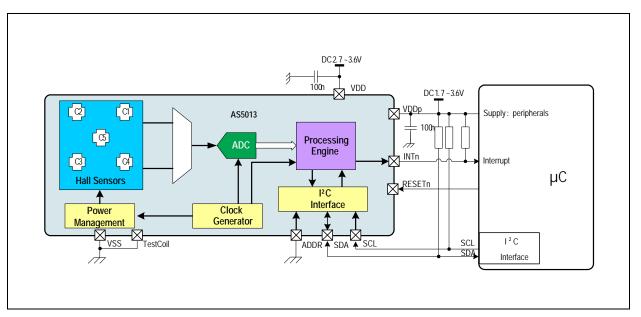
The AS5013 requires only a few external components in order to operate immediately when connected to the host microcontroller.

Only 4 wires are needed for a simple application using a single power supply: two wires for power and two wires for the I<sup>2</sup>C communication. A fifth connection can be added in order to send an interrupt to the host CPU when the magnet is moving away from the center and to inform that a new valid coordinate can be read.

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Figure 9: Electrical Connection of AS5013 with Microcontroller



### **XY Coordinates Interpretation**

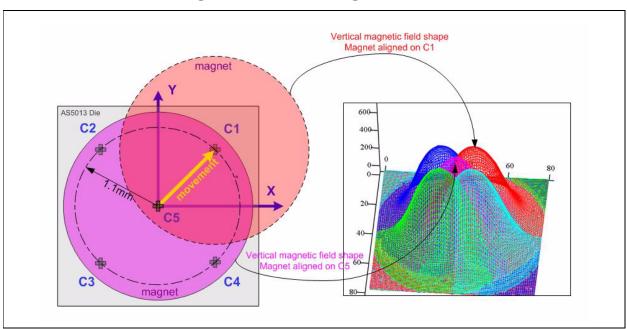
The movement of the magnet over the Hall elements causes response which is geometrically distributed like a bell-shaped curve.

The maximum magnet travel is a circle of 2mm radius around the center of the AS5013. The Hall elements C1..C4 are placed on a circle centered on the middle of the package. The Hall element C5, placed exactly in the middle is used for better linearity response with magnet displacement larger than ±1.0mm.

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Figure 10:
Hall Element Placement and Magnetic Field when the Magnet Is Centered Over Each Hall Element



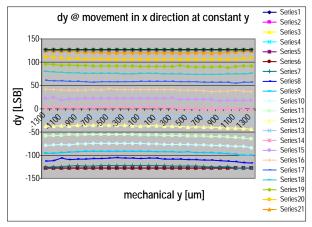
#### **Transfer Function**

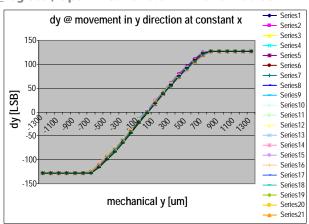
AS5013 has the possibility to adjust the transfer function for the used magnet and a specific range to optimize the linearity and resolution. The value will be provided from ams and has to be written in the algorithm related registers M\_ctrl [0x2B], J\_ctrl [0x2C], T\_ctrl [0x2D] during the initialization phase.

Please contact ams for parameter settings.

Below is the optimal setup for a range of  $\pm 0.6$  mm to obtain the best dynamic range from XY registers -128~+127 with one given magnet airgap, with d2x0.8mm axial magnet.

Figure 11: Example of Transfer Function Y\_displacement vs. Y\_register, Optimized for 0.6mm Travel Radius





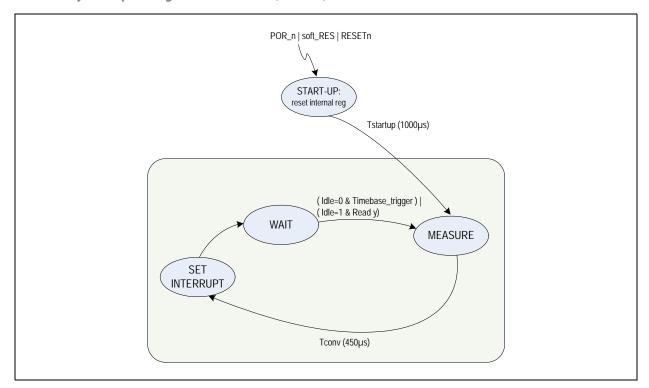
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#### **Power Modes**

The AS5013 can operate in two different power modes, depending on the power consumption requirements of the whole system.

Figure 12:
Readout Cycle Depending on Power Mode (idle bit)



#### START-UP

After power up and after applying a soft reset (Reg 0Fh [1]) or hardware reset (RESETn input, LOW pulse >100ns), AS5013 enters the START-UP state. During this state the internal registers are loaded with their reset values. After min. Tstartup =  $1000\mu s$ , the AS5013 will perform one measurement and switches automatically into the WAIT state.

#### **MEASURE**

The Hall element data are measured, x/y coordinates are calculated and available in registers 10h and 11h after Tconv =  $450\mu s$  max.

#### **SET INTERRUPT**

The INTn output is set, depending on the interrupt mode configured in the control register Reg 0Fh [2] and Reg 0Fh [3]

#### WAIT

The module is now in waiting status. A new measurement will occur depending on the power mode (Reg 0Fh [7] Idle = 0 or 1) and the Timebase Reg 0Fh [6:4]

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### I<sup>2</sup>C Interface

The AS5013 supports the 2-wire high-speed I<sup>2</sup>C protocol in device mode, according to the NXP specification UM10204.

The host MCU (master) has to initiate the data transfers. The 7-bit device address of the AS5013 depends on the state at the pin ADDR.

ADDR =  $0 \rightarrow \text{Slave address} = '100\ 0000'\ (40h)$ 

ADDR =  $1 \rightarrow \text{Slave address} = '100\ 0001'\ (41h)$ 

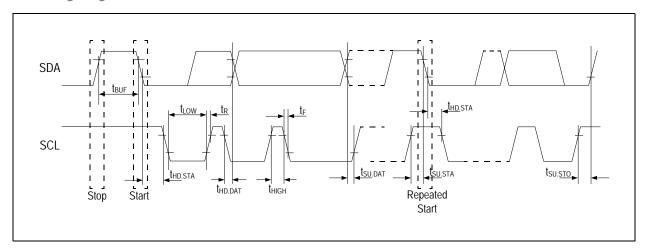
For other I<sup>2</sup>C addresses, please contact **ams**.

Supported modes (slave mode):

- Random/Sequential Read
- Byte/Page Write
- Standard Mode: 0 to 100kHz clock frequency
- Fast Mode: 0 to 400kHz clock frequency
- High Speed: 0 to 3.4MHz clock frequency

The SDA signal is bidirectional and is used to read and write the serial data. The SCL signal is the clock generated by the host MCU, to synchronize the SDA data in read and write mode. The maximum I<sup>2</sup>C clock frequency is 3.4MHz, data are triggered on the rising edge of SCL.

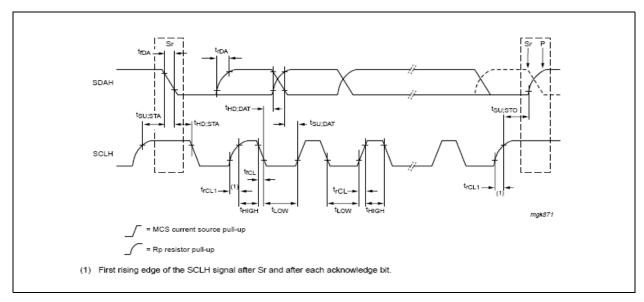
Figure 13: I<sup>2</sup>C Timing Diagram for FS-Mode



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Figure 14: Timing Diagram for HS-Mode



# I<sup>2</sup>C Electrical Specification

Standard-mode, Fast-mode, High speed-mode

Figure 15: I<sup>2</sup>C Electrical Specification

Symbol	Parameter	Condition	Min	Max	Units
V <sub>IL</sub>	LOW-level input voltage		-0.5	0.3VDDp	V
V <sub>IH</sub>	HIGH-level input voltage		0.7VDDp	VDDp + 0.5 <sup>(1)</sup>	V
V <sub>hys</sub>	Hysteresis of Schmitt Trigger inputs	VDDp < 2V	0.1VDDp		V
V <sub>OL</sub>	LOW-level output voltage (open-drain or open-collector) at 3mA sink current	VDDp < 2V	-	0.2VDDp	V
l <sub>OL</sub>	LOW-level output current	V <sub>OL</sub> = 0.4V	-	-	mA
I <sub>CS</sub>	Pull-up current of SCLH current source	SCLH output levels between 0.3VDDp and 0.7VDDp	3	12	mA
t <sub>SP</sub>	Pulse width of spikes that must be suppressed by the	In HS-mode	-	10 <sup>(2)</sup>	ns
,24	input filter	In Fast-mode		50 <sup>(2)</sup>	ns
l <sub>i</sub>	Input current at each I/O Pin	Input voltage between 0.1VDDp and 0.9VDDp		10 <sup>(3)</sup>	μΑ

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Symbol	Parameter	Condition	Min	Max	Units
C <sub>B</sub>	Total capacitive load for each bus line		-	400	pF
C <sub>I/O</sub>	I/O capacitance (SDA, SCL)		-	10	pF

#### Note(s):

- 1. Maximum VIH = VDDpmax +0.5V or 5.5V, which ever is lower.
- 2. Input filters on the SDA and SCL inputs suppress noise spikes of less than 50ns in Fast-mode and 10ns in HS-mode.
- 3. I/O pins of Fast-mode and Fast-mode plus devices must not obstruct the SDA and SCL lines if VDDp is switched off.

# Figure 16: I<sup>2</sup>C Timing

Symbol	Parameter Condition		Fast-Mode		HS-Mode C <sub>B</sub> =100pF		HS-Mode C <sub>B</sub> =400pF <sup>(1)</sup>		Unit
			Min	Max	Min	Max	Min	Max	
f <sub>SCLK</sub>	SCL clock Frequency		-	400	-	3400	-	1700	kHz
t <sub>BUF</sub>	Bus Free Time; time between STOP and START Condition		500	-	500	-	500	-	ns
t <sub>HD;STA</sub>	Hold Time; (Repeated) START Condition <sup>(2)</sup>		600	-	160	-	160	-	ns
t <sub>LOW</sub>	LOW Period of SCL Clock		1300	-	160	-	320	-	ns
t <sub>HIGH</sub>	HIGH Period of SCL Clock		600	-	60	-	120	-	ns
t <sub>SU;STA</sub>	Setup Time for a Repeated START condition		600	-	160	-	160	-	ns
t <sub>HD;DAT</sub>	Data Hold Time <sup>(3)</sup>		0	900	0	70	0	150	ns
t <sub>SU;DAT</sub>	Data Setup Time <sup>(4)</sup>		100	-	10	-	10	-	ns

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Symbol	Parameter	Condition	Fast-M	lode	HS-M C <sub>B</sub> =1		HS-N C <sub>B</sub> =40	Mode 00pF <sup>(1)</sup>	Unit
			Min	Max	Min	Max	Min	Max	
t <sub>rCL</sub>	Rise time of SCLH signal	External pull-up source of 3mA	-	-	10	40	20	80	ns
t <sub>rCL1</sub>	Rise time of SCLH signal after repeated START condition and after an acknowledg e bit	External pull-up source of 3mA	-	-	10	80	20	160	ns
t <sub>R</sub>	Rise Time of SDA and SCL Signals		20+0.1C <sub>B</sub>	120	-	-	-	-	ns
t <sub>F</sub>	Fall time of SDA and SCL signals		20+0.1C <sub>B</sub>	120	-	-	-	-	ns
t <sub>SU;STO</sub>	Setup Time for STOP Condition		600	-	160	-	160	-	ns
V <sub>nL</sub>	Noise margin at LOW level	For each connected device	0.1VDDp	-	0.1 VDDp	-	0.1 VDDp	-	<b>V</b>
V <sub>nH</sub>	Noise margin at HIGH level	(including hysteresis)	0.2VDDp	-	0.2 VDDp	-	0.2 VDDp	-	V

#### Note(s):

- 1. For bus line loads CB between 100pF and 400 pF the timing parameters must be linearly interpolated.
- 2. After this time the first clock is generated.
- 3. A device must internally provide a minimum hold time (300n for Fast-mode, 80ns / max 150ns for High-speed mode) for the SDA signal (referred to the VIHmin of the SCL) to bridge the undefined region of the falling edge of SCL.
- 4. A fast-mode device can be used in standard-mode system, but the requirement tSU;DAT = 250ns must then be met. This is automatically the case if the device does not stretch the LOW period of the SCL signal. If such a device does stretch the LOW period of the SCL signal, it must output the next data bit to the SDA line tRmax + tSU;DAT = 1000 + 250 = 1250ns before the SCL line is released.

#### I<sup>2</sup>C Modes

The AS5013 supports the I<sup>2</sup>C bus protocol. A device that sends data onto the bus is defined as a transmitter and a device receiving data as a receiver. The device that controls the message is called a master. The devices that are controlled by the master are referred to as slaves. A master device that generates the serial clock (SCL), controls the bus access, and generates the START and STOP conditions must control the bus.

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The AS5013 operates as a slave on the I<sup>2</sup>C bus. Connections to the bus are made through the open-drain I/O lines SDA and the input SCL. Clock stretching is not included.

#### **Automatic Increment of Address Pointer**

The AS5013 slave automatically increments the address pointer after each byte transferred. The increase of the address pointer is independent from the address being valid or not.

#### **Invalid Addresses**

If the user sets the address pointer to an invalid address, the address byte is not acknowledged. Nevertheless a read or write cycle is possible. The address pointer is increased after each byte.

#### Reading

When reading from a wrong address, the AS5013 slave data returns all zero. The address pointer is increased after each byte. Sequential read over the whole address range is possible including address overflow.

#### Writing

A write to a wrong address is not acknowledged by the AS5013 slave, although the address pointer is increased. When the address pointer points to a valid address again, a successful write access is acknowledged. Page write over the whole address range is possible including address overflow.

The following bus protocol has been defined:

- Data transfer may be initiated only when the bus is not busy.
- During data transfer, the data line must remain stable whenever the clock line is HIGH. Changes in the data line while the clock line is HIGH are interpreted as start or stop signals.

Accordingly, the following bus conditions have been defined:

### **Bus Not Busy:**

Both data and clock lines remain HIGH.

#### **Start Data Transfer:**

A change in the state of the data line, from HIGH to LOW, while the clock is HIGH, defines a START condition.

#### **Stop Data Transfer:**

A change in the state of the data line, from LOW to HIGH, while the clock line is HIGH, defines the STOP condition.

#### **Data Valid:**

The state of the data line represents valid data when, after a START condition, the data line is stable for the duration of the HIGH period of the clock signal. The data on the line must be changed during the LOW period of the clock signal. There is one clock pulse per bit of data. Each data transfer is initiated with a

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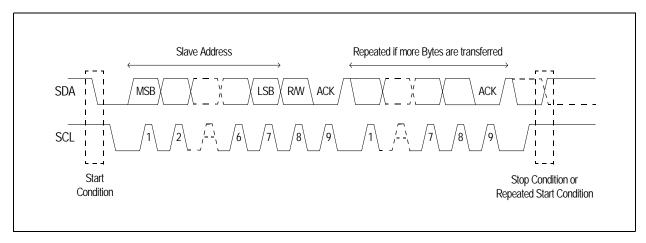
START condition and terminated with a STOP condition. The number of data bytes transferred between START and STOP conditions are not limited, and are determined by the master device. The information is transferred byte-wise and each receiver acknowledges with a ninth bit.

#### Acknowledge:

Each receiving device, when addressed, is obliged to generate an acknowledge after the reception of each byte. The master device must generate an extra clock pulse that is associated with this acknowledge bit.

A device that acknowledges must pull down the SDA line during the acknowledge clock pulse in such a way that the SDA line is stable LOW during the HIGH period of the acknowledge-related clock pulse. Of course, setup and hold times must be taken into account. A master must signal an end of READ access to the slave by not generating an acknowledge bit on the last byte that has been clocked out of the slave. In this case, the slave must leave the data line HIGH to enable the master to generate the STOP condition.

Figure 17:
Data Read (Write Pointer, Then Read) - Slave Receive and Transmit



Depending upon the state of the R/W bit, two types of data transfer are possible:

- Data transfer from a master transmitter to a slave receiver: The first byte transmitted by the master is the slave address, followed by R/W = 0. Next follows a number of data bytes. The slave returns an acknowledge bit after each received byte. If the slave does not understand the command or data it sends a "not acknowledge". Data is transferred with the most significant bit (MSB) first.
- Data transfer from a slave transmitter to a master receiver: The master transmits the first byte (the slave address). The slave then returns an acknowledge bit, followed by the slave transmitting a number of data bytes. The master returns an acknowledge bit after all received bytes other than the last byte. At the end of the last received byte, a "not acknowledge" is returned. The master device generates all of the serial clock pulses and

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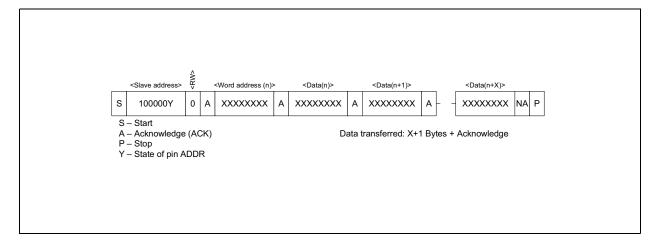


the START and STOP conditions. A transfer is ended with a STOP condition or with a repeated START condition. Since a repeated START condition is also the beginning of the next serial transfer, the bus is not released. Data is transferred with the most significant bit (MSB) first.

The AS5013 can operate in the following two modes:

• Slave Receiver Mode (Write Mode): Serial data and clock are received through SDA and SCL. Each byte is followed by an acknowledge bit (or by a not acknowledge depending on the address-pointer pointing to a valid position). START and STOP conditions are recognized as the beginning and end of a serial transfer. Address recognition is performed by hardware after reception of the slave address and direction bit (see Figure 18). The slave address byte is the first byte received after the START condition. The slave address byte contains the 7-bit AS5013 address, which is stored in the OTP memory. The 7-bit slave address is followed by the direction bit (R/W), which, for a write, is 0. After receiving and decoding the slave address byte the device outputs an acknowledge on the SDA. After the AS5013 acknowledges the slave address + write bit, the master transmits a register address to the AS5013. This sets the address pointer on the AS5013. If the address is a valid readable address the AS5013 answers by sending an acknowledge. If the address-pointer points to an invalid position a "not acknowledge" is sent. The master may then transmit zero or more bytes of data. In case of the address pointer pointing to an invalid address the received data are not stored. The address pointer will increment after each byte transferred independent from the address being valid. If the address-pointer reaches a valid position again, the AS5013 answers with an acknowledge and stores the data. The master generates a STOP condition to terminate the data write.

Figure 18: Data Write - Slave Receiver Mode



• Slave Transmitter Mode (Read Mode): The first byte is received and handled as in the slave receiver mode. However, in this mode, the direction bit indicates that the transfer direction is reversed. Serial data is transmitted on

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SDA by the AS5013 while the serial clock is input on SCL. START and STOP conditions are recognized as the beginning and end of a serial transfer. The slave address byte is the first byte received after the master generates a START condition. The slave address byte contains the 7-bit AS5013 address. The default address is 40h. The 7-bit slave address is followed by the direction bit (R/W), which, for a read, is 1. After receiving and decoding the slave address byte the device outputs an acknowledge on the SDA line. The AS5013 then begins to transmit data starting with the register address pointed to by the register pointer. If the register pointer is not written to before the initiation of a read mode the first address that is read is the last one stored in the register pointer. The AS5013 must receive a "not acknowledge" to end a read.

Figure 19:
Data Read (from Current Pointer Location) - Slave Transmitter Mode

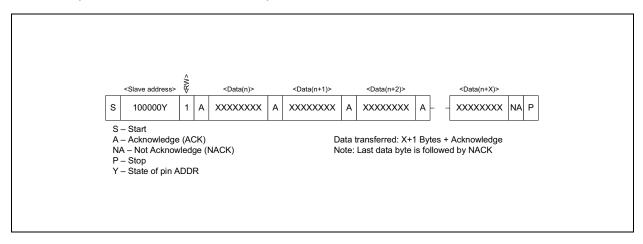
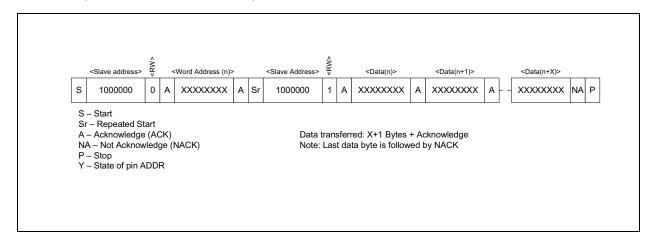


Figure 20:
Data Read (from New Pointer Location) - Slave Transmitter Mode



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#### High Speed Mode

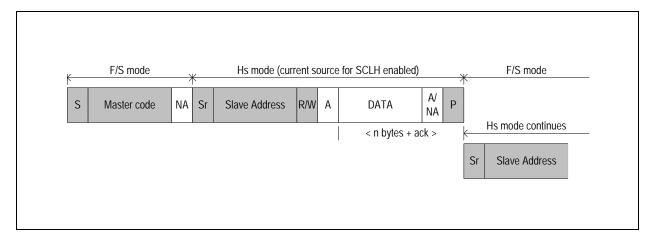
The AS5013 is capable to work in HS-mode.

For switching to HS-mode the Master has to send the sequence: START, MASTER CODE, NACK. This sequence is sent in FS-mode. As no device is allowed to acknowledge the master code, the master code is followed by a not-acknowledge. After a device receives the master code it has to switch from FS-settings to HS-settings within tSU.STA which is 160ns for HS-mode. The device stays in HS-mode as long as it does not receive a STOP command. After receiving a STOP command it has to switch back form HS-settings to FS-settings, which has to be competed within the minimum bus free time  $t_{\rm BUF}$  which is 500ns.

When switching to HS-mode the slave has to:

- Adapt the SDAH and SCLH input filters according to the spike suppression requirement required in HS-mode. In HS-mode spikes up to 10ns, in FS-mode spikes up to 50ns have to be suppressed.
- Adapt the setup and hold times according to the HS-mode requirement. In HS-mode an internal hold time for SDA for START/STOP detection of 80ns (max. 150ns), in FS-mode an internal hold time of 160ns (max. 250ns) has to be provided.
- Adapt the slope control for SDAH output stage.

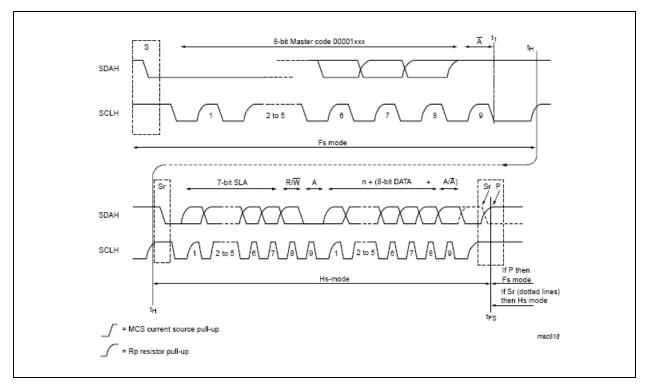
Figure 21:
Data Transfer Format in HS-Mode



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Figure 22: A Complete HS-Mode Transfer



#### **Automatic Increment of Address Pointer**

The AS5013 slave automatically increments the address pointer after each byte transferred. The increase of the address pointer is independent from the address being valid or not.

#### **Invalid Addresses**

If the user sets the address pointer to an invalid address, the address byte is not acknowledged. Nevertheless a read or write cycle is possible. The address pointer is increased after each byte.

#### Reading:

When reading from a wrong address, the AS5013 slave returns all zero. The address pointer is increased after each byte. Sequential read over the whole address range is possible including address overflow.

#### Writing:

A write to a wrong address is not acknowledged by the AS5013 slave, although the address pointer is increased. When the address pointer points to a valid address again, a successful write accessed is acknowledged. Page write over the whole address range is possible including address overflow.

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# SDA, SCL Input Filters

Input filters for SDA and SCL inputs are included to suppress noise spikes of less than 50ns. Furthermore, the SDA line is delayed by 120ns to provide an internal hold time for Start/Stop detection to bridge the undefined region of the falling edge of SCL. The delay needs to be smaller than  $t_{\mbox{\scriptsize HD.STA}}$  260ns.

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# I<sup>2</sup>C Registers

Figure 23: Control Register 1 (0Fh)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Idle	Time base bit[2]	Time base bit[1]	Time base bit[0]	INT_ disable	INT_ function	Soft_rst	Data_ valid
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R
Reset value:	1111 0000						

Bit	Bit Description
	0= Low Power Mode The measurements are triggered with an internal low power oscillator – the user can select between 8 different timings by setting the low power timebase (Control Register 1 [6:4])
7	1=Idle Mode (default) A new measurement cycle is started after the I <sup>2</sup> C ACK bit following the read out of the Y_res_int register 11h. The readout rate and thus the power consumption is externally controlled by the host MCU.
6:4	Configure the time base of the automatic wakeup in Low Power Mode (see Figure 24).
3	0=Interrupt output INTn is enabled (default)
3	1=Interrupt output INTn is disabled and is fixed to Hi-Z
	0=Interrupt output INTn is active '0' after each measurement (default):  - Automatically triggered in Low Power mode, depending on the time base chosen  - 450µs after Y readout in Idle mode  The interrupt is cleared by the I <sup>2</sup> C ACK bit after reading the Y_res_int 11h. In block read mode, the several other bytes could be transferred before the interrupt is cleared.
2	1=Interrupt output INTn is active '0' when the movement of the magnet exceeds the Dead Zone area. (see Figure 25).  The Dead Zone area is set by registers Xp (Reg 12h), Xn (Reg 13h), Yp (Reg 14h), Yn (Reg 15h).  The interrupt is cleared by the I <sup>2</sup> C ACK bit after reading the Y_res_int register 11h, and will be active '0' at the next measurement if the magnet is still in the Detection Area. In block read mode, several other bytes could be transferred before the interrupt is cleared when the Y_res_int register is read.  It is recommended to use this mode with the Low Power mode (Idle = 0), in order to wake up automatically a system when the magnet has been moved away from the center. The polling time is set by the Low Power time base bit [6:4].
	0=Normal mode (default)
1	1=Reset mode. All the internal registers are loaded with their reset value. The Control Register 1 is loaded as well with the value 1111 0000, then the Soft_rst bit goes back to 0 (Normal mode) once the internal reset sequence is finished.

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Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	0=Conversion of new coordinates ongoing, no valid coordinate is present in the X and Y_res_int registers. Reading those registers at that moment can give wrong values.						
	1=New coordin	ate values are rea	idy in X and Y_res_	_int register:	s.		

#### Note(s):

1. The values in Control Register 1, X\_register and Y\_res\_int register are frozen when the I<sup>2</sup>C address pointer is set to 0Fh, 10h or 11h. This ensures that the Data\_valid bit, X and Y values are taken at the same time. In order to get updated values from those registers, set the address pointer to any other address.

Figure 24: Configuration

Low Power Time Base CONFIG_REG1 0Fh [6:4]	$\Delta  extsf{t}_{ extsf{timebase}}$ (ms)	Average Core Current IDD (μA) @T <sub>AMB</sub> = 25°C
000b	20	190
001b	40	97
010b	80	50
011b	100	40
100b	140	30
101b	200	22
110b	260	17
111b (default)	320	15

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Figure 25:
Dead Zone Representation with INT\_function=1

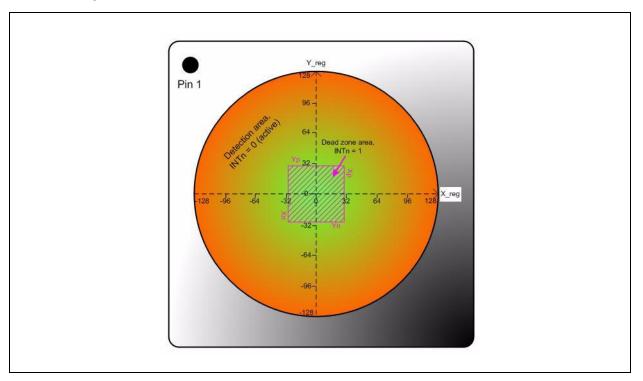


Figure 26: X Register (10h)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
X[7]	X[6]	X[5]	X[4]	X[3]	X[2]	X[1]	X[0]
R	R	R	R	R	R	R	R
Reset value:	0000 0000						

Bit	Bit Description
7:0	X coordinate, Two's complement format (signed -128 ~ +127).

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Figure 27:

Y\_res\_int Register (11h)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Y[7]	Y[6]	Y[5]	Y[4]	Y[3]	Y[2]	Y[1]	Y[0]
R	R	R	R	R	R	R	R
Reset value:	0000 0000				•		

Bit	Bit Description
7:0	Y coordinate, Two's complement format (signed -128 ~ +127).  Reading this register will reset the INTn output to Hi-Z after the ACK bit of Y_res_int register readback.

Figure 28: Xp Register (12h)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Xp[7]	Xp[6]	Xp[5]	Xp[4]	Xp[3]	Xp[2]	Xp[1]	Xp[0]
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value:	0000 0101 (5d)			•	•		

Bit	Bit Description
7:0	Xp range value, Two's complement (signed: -128 ~ +127).  Determines the LEFT threshold for the activation of INTn output (if output enabled), when bit INT_function = 1 (See Figure 23).

# Figure 29: Xn Register (13h)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Xn[7]	Xn[6]	Xn[5]	Xn[4]	Xn[3]	Xn[2]	Xn[1]	Xn[0]
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value:	1111 1011 (-5d)						

Bit	Bit Description
7:0	Xn range value, Two's complement (signed: -128 $\sim$ +127). Determines the RIGHT threshold for the activation of INTn output (if output enabled), when bit INT_function = 1 (See Figure 23).

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# Figure 30: Yp Register (14h)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Yp[7]	Yp[6]	Yp[5]	Yp[4]	Yp[3]	Yp[2]	Yp[1]	Yp[0]
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value: 0000 0101 (5d)							

E	Bit	Bit Description
7	7:0	Yp range value, Two's complement (signed: $-128 \sim +127$ ). Determines the TOP threshold for the activation of INTn output (if output enabled), when bit INT_function = 1 (See Figure 23).

Figure 31: Yn Register (15h)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Yn[7]	Yn[6]	Yn[5]	Yn[4]	Yn[3]	Yn[2]	Yn[1]	Yn[0]
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value: 1111 1011 (-5d)							

Bit	Bit Description
7:0	Yn range value, Two's complement (signed: -128 ~ +127).  Determines the BOTTOM threshold for the activation of INTn output (if output enabled), when bit INT_function = 1 (See Figure 23).

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Figure 32:

M\_ctrl Register (2Bh)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
M_ctrl[7]	M_ctrl[6]	M_ctrl[5]	M_ctrl[4]	M_ctrl[3]	M_ctrl[2]	M_ctrl[1]	M_ctrl[0]
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value: 0000 0000 (00h)							

Bit	Bit Description			
7:0	Middle Hall element C5 control register to improve the linearity of XY outputs for the whole mechanical XY displacement of the magnet. Use the default value for d=2*0.8mm standard axial magnet.  For more information on how to configure this parameter, please contact <b>ams</b> .			

Figure 33: J\_ctrl Register (2Ch)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
J_ctrl[7]	J_ctrl[6]	J_ctrl[5]	J_ctrl[4]	J_ctrl[3]	J_ctrl[2]	J_ctrl[1]	J_ctrl[0]
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value: 0000 0110 (06h)							

Bit	Bit Description			
7:0	Sector dependent attenuation of the outer Hall elements C1C4 in order to improve the linearity of XY outputs for the whole mechanical XY displacement of the magnet. Use the default value for $d=2*0.8$ mm standard axial magnet. For more information on how to configure this parameter, please contact <b>ams</b> .			

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Figure 34: T\_ctrl Register (2Dh)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
T_ctrl[7]	T_ctrl[6]	T_ctrl[5]	T_ctrl[4]	T_ctrl[3]	T_ctrl[2]	T_ctrl[1]	T_ctrl[0]
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value: 0000 1001 (09h)							

Bit	Bit Description
7:0	Scaling control register.  This register controls the scaling factor of the XY coordinates to fit to the 8-bit X and Y_res_int register (full dynamic range). The following table includes scaling factors referenced to the default setting T_ctrl = 9 (100% scaling).

T_ctrl	Scaling Factor %
31	31.3
30	32.2
29	33.4
28	34.6
27	35.7
26	37.1
25	38.5
24	40.0
23	41.6
22	43.6
21	45.5
20	47.7
19	50.0
18	52.5
17	55.5
16	58.8
15	62.5
14	66.6
13	71.5

T_ctrl	Scaling Factor %
47	117.6
7	125.0
45	133.4
6	142.8
43	153.9
5	166.6
41	181.8
4	200.0
79	210.5
39	222.3
77	235.4
3	250.0
75	266.6
37	285.7
73	307.6
2	333.4
71	363.7
35	400.0
69	444.5



T_ctrl	Scaling Factor %
12	77.0
11	83.4
10	90.8
9	100.0
8	111.1

T_ctrl	Scaling Factor %
1	500.0
67	571.5

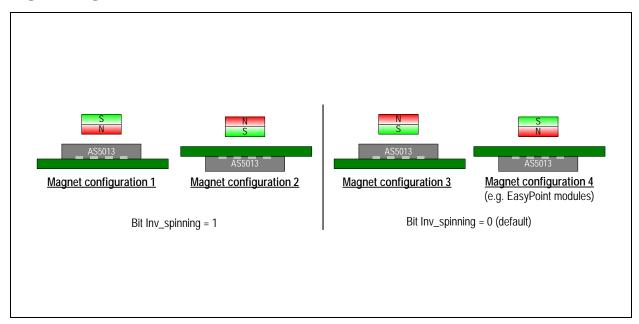
Figure 35: Control Register 2 (2Eh)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Test	Test	Test	ext_clk_en	use_static_ offset	EN_offset_ comp	inv_ spinning	Test
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value	e: 1000 0100						

Bit	Bit Description
7	Test bit. Must configured '1'.
6:4	Test bit. Must configured '000'.
3	Test bit. Must configured '0'.
2	Test bit. Must configured '1'.
1	Magnet Polarity bit. Must be set after power up, depending on how the magnet is placed Figure 36
0	Test bit. Must configured '0'.



Figure 36: **Magnet Configuration** 



**Note:** In order to know the polarity of the magnet without any testing device, please refer to Registers Initialization

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### Hall Element Direct Read Registers (16h to 29h)

Each Hall element C1..C5 can be read independently, after each interrupt (data ready).

One hall element value consists of two 12-bit signed-registers: Cx\_neg and Cx\_pos. For each conversion cycle (i.e. after a readout or Y\_res in idle mode, or at each time-based conversion cycle in Low Power mode), each Hall element is read twice: With normal spin (result Cx\_pos) and then with inverted spin (result Cx\_neg) in order to remove any Hall voltage offset from the Hall elements.

The formula to read any Hall element Cx:

Figure 37: Hall Element Direct Read Registers (2Ah)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	0	AGC5	AGC4	AGC3	AGC2	AGC1	AGC0
R	R	R/W	R/W	R/W	R/W	R/W	R/W
Reset value: 0010 0000							

The AGC register controls the sensitivity of each Hall element C1...C5, in order to stay in the larger dynamic range of the 12-bit ADC of the AS5013. In order to determine the best value to be set during the AS5013 initialization, place your magnet on the 0,0 position (centered on C5 Hall element), and increase the AGC value to obtain the nearest value to 2867 (= 70% of 4096).

It is possible that this value cannot be reached with small magnets or with large airgaps. In that case set AGC to 3Fh, which is the maximum sensitivity.

#### Power ON

The AS5013 has a Power ON Reset (POR) cell to monitor the VDD voltage at startup and reset all the internal registers.

After the internal reset is completed, the POR cell is disabled in order to save current during normal operation.

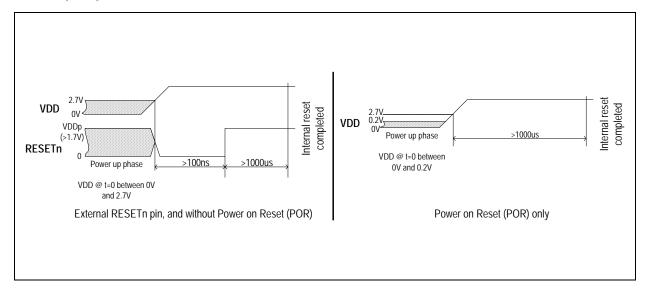
If VDD drops below 2.7V down to 0.2V, the POR cell will not be enabled back, and the registers will not be correctly reseted or can get random values.

**Note:** It is highly recommended to control the external RESETn signal by applying a LOW pulse of >100ns once VDD reached 2.7V and VDDp reached 1.7V.

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Figure 38: Power-Up Sequence



# **Registers Initialization**

After Power Up, the following sequence must be performed:

- 1. VDD and VDDp Power up, and reached their nominal values (VDD>2.7V, VDDp>1.7V).
- 2. RESETn LOW during >100ns
- 3. Delay 1000µs
- 4. Loop check register [0Fh] until the value F0h or F1h is present (reset finished, registers to their default values)
- Optional: Write value 86h into register [2Eh] → Invert magnet polarity. See Control Register 2 (2Eh) on page 22.
- 6. Configure register [2Bh] → Configure M\_ctrl middle Hall element control
- 7. Configure register [2Ch] → Configure J\_ctrl attenuation factor
- 8. Configure register [2Dh] → Configure T\_ctrl scaling factor
- Configure the wanted Power Mode into register [0Fh] (Idle mode or Low Power Mode with Timebase configuration)
- 10. X Y coordinates are ready to be read.

**Note:** In order to detect if the magnet polarity is correct, read the C5 middle Hall element when the magnet is centered.

$$C5 = (C5_{pos} - C5_{neg}) / 2$$

C5 must always be positive.

If C5 is negative, then invert the bit inv\_spinning in the Control Register 2 (2Eh). C5 will become positive.

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# **Registers Table**

The following registers / functions are accessible over the serial  $I^2C$  interface.

Figure 39: Registers

Register	Number of Bits	Access	Address	Format	Reset Value	Bit	Description
IC Identification							
ID Code	8	R	0C		0Ch	<7:0>	8-bit Manufacture ID Code
ID Version	8	R	0D		0Dh	<7:0>	8-bit Component ID Version
Silicon Revision	8	R	0E		00h	<7:0>	8-bit Silicon Revision
			Control_	register_1			
Idle	1	R/W	0Fh		1b	<7>	1: Idle mode 0: Low Power mode
Low_ power_ timebase	3	R/W	0Fh		111b	<6:4>	Low Power readout time base register
INT_ disable	1	R/W	0Fh		0b	<3>	Disables the interrupt functionality. 1: Interrupt disabled 0: Interrupt enabled
INT_ function	1	R/W	0Fh		0b	<2>	Interrupt control register 0: Interrupt goes low with every new calculated x/y coordinates 1: Interrupt pin goes low in when new x/y coordinates are calculated and the magnet has exited the xp, xn, yp, yn threshold values
soft_rst	1	R/W	0Fh		0b	<1>	Soft Reset 0: Normal mode 1: All registers return to their respective reset value



Register	Number of Bits	Access	Address	Format	Reset Value	Bit	Description
data_valid	1	R	0Fh		0b	<0>	Data valid indicator 0: X/Y calculation ongoing 1: X/Y calculation finished, coordinates ready
			X/Y Coordi	nate Registe	er		
х	8	R	10h	two's comp.	00h	<7:0>	Result
y_res_int	8	R	11h	two's comp.	00h	<7:0>	Result, resets the interrupt flag at the value ACK
			Range	Settings			
хр	8	R/W	12h	two's comp.	5h (5 dec)	<7:0>	Wake up threshold @ positive X-direction
xn	8	R/W	13h	two's comp.	FBh (-5 dec)	<7:0>	Wake up threshold @ negative X-direction
ур	8	R/W	14h	two's comp.	5h (5 dec)	<7:0>	Wake up threshold @ positive Y-direction
yn	8	R/W	15h	two's comp.	FBh (-5 dec)	<7:0>	Wake up threshold @ negative Y-direction
			Channel \	/oltages (3)			
c4_neg <11:8>	4	R	16h	two's comp.	00h	<3:0> <7:4>	Voltage @ channel 4, negative current spinning Sign extended to 8-bit
c4_neg <7:0>	8	R	17h	two's comp.	00h	<7:0>	Voltage @ channel 4, negative current spinning
c4_pos <11:8>	4	R	18h	two's comp.	00h	<3:0> <7:4>	Voltage @ channel 4, positive current spinning Sign extended to 8-bit
c4_pos <7:0>	8	R	19h	two's comp.	00h	<7:0>	Voltage @ channel 4, positive current spinning

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Register	Number of Bits	Access	Address	Format	Reset Value	Bit	Description
c3_neg <11:8>	4	R	1Ah	two's comp.	00h	<3:0> <7:4>	Voltage @ channel 3, negative current spinning Sign extended to 8-bit
c3_neg <7:0>	8	R	1Bh	two's comp.	00h	<7:0>	Voltage @ channel 3, negative current spinning
c3_pos <11:8>	4	R	1Ch	two's comp.	00h	<3:0> <7:4>	Voltage @ channel 3, positive current spinning Sign extended to 8-bit
c3_pos <7:0>	8	R	1Dh	two's comp.	00h	<7:0>	Voltage @ channel 3, positive current spinning
	I	l.	I	I			
c2_neg <11:8>	4	R	1Eh	two's comp.	00h	<3:0> <7:4>	Voltage @ channel 2, negative current spinning Sign extended to 8-bit
c2_neg <7:0>	8	R	1Fh	two's comp.	00h	<7:0>	Voltage @ channel 2, negative current spinning
c2_pos <11:8>	4	R	20h	two's comp.	00h	<3:0> <7:4>	Voltage @ channel 2, positive current spinning Sign extended to 8-bit
c2_pos <7:0>	8	R	21h	two's comp.	00h	<7:0>	Voltage @ channel 2, positive current spinning
c1_neg <11:8>	4	R	22h	two's comp.	00h	<3:0> <7:4>	Voltage @ channel 1, negative current spinning Sign extended to 8-bit
c1_neg <7:0>	8	R	23h	two's comp.	00h	<7:0>	Voltage @ channel 1, negative current spinning



Register	Number of Bits	Access	Address	Format	Reset Value	Bit	Description
c1_pos <11:8>	4	R	24h	two's comp.	00h	<3:0> <7:4>	Voltage @ channel 1, positive current spinning Sign extended to 8-bit
c1_pos <7:0>	8	R	25h	two's comp.	00h	<7:0>	Voltage @ channel 1, positive current spinning
c5_neg <11:8>	4	R	26h	two's comp.	00h	<3:0> <7:4>	Voltage @ channel 5, negative current spinning Sign extended to 8-bit
c5_neg <7:0>	8	R	27h	two's comp.	00h	<7:0>	Voltage @ channel 5, negative current spinning
c5_pos <11:8>	4	R	28h	two's comp.	00h	<3:0> <7:4>	Voltage @ channel 5, positive current spinning Sign extended to 8-bit
c5_pos <7:0>	8	R	29h	two's comp.	00h	<7:0>	Voltage @ channel 5, positive current spinning
Hall Bias Currents							
AGC	8	RW	2Ah		00b 20h	<7:6> <5:0>	Not implemented (read 00b) 6-bit AGC value (if an AGC algorithm implemented in the $\mu$ C)

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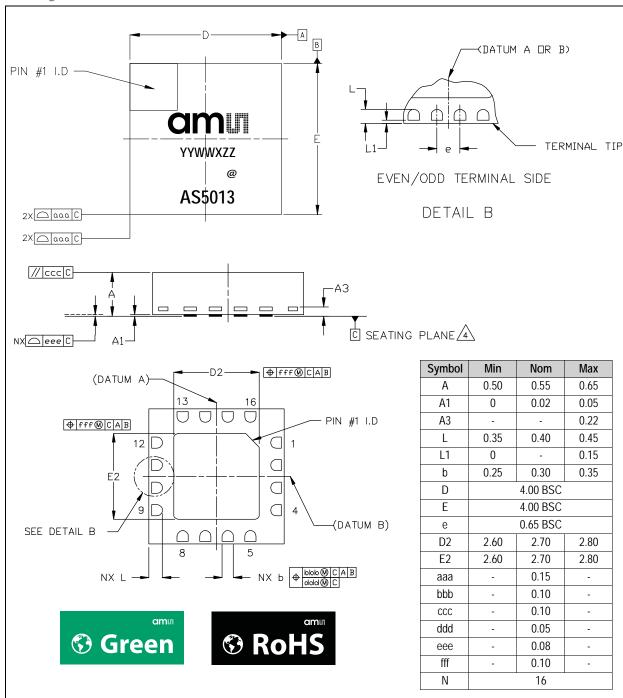
Dovietov	Number	A	A dalue e e	Farmant	Reset	D:4	Decembrish	
Register	of Bits	Access	Address	Format	Value	Bit	Description	
Control Register for the Algorithm								
M_ctrl	8	R/W	2Bh		00h	<7:0>	Control register for the middle Hall element C5. If the register is zero the middle Hall element is not used for the XY calculation	
J_ctrl	8	R/W	2Ch		06h	<7:0>	Control register for the sector dependent attenuation of the outer Hall elements	
T_ctrl	8	R/W	2Dh		09h	<7:0>	Scale input to fit to the 8-bit result register	
			Control_	register_2				
Test	1	R/W	2Eh		1b	<7>	Test only, must be '1'	
Test	1	R/W	2Eh		0b	<6>	Test only, must be '0'	
Test	1	R/W	2Eh		0b	<5>	Test only, must be '0'	
ext_clk_en	1	R/W	2Eh		0b	<4>	Test only, must be '0'	
use_ static_ offset	1	R/W	2Eh		0b	<3>	Test only, must be '0'	
EN_offset_ comp	1	R/W	2Eh		1b	<2>	Test only, must be '1'	
inv_ spinning	1	R/W	2Eh		0b	<1>	Invert the channel voltage. Set to invert the magnet polarity	
pptrim_en	1	R/W	2Eh		0b	<0>	Factory only, must be '0'	



# **Package Drawings & Markings**

The device is available in a 16-pin QFN (4mmx4mmx0.55mm) package.

Figure 40: Drawings and Dimensions



#### Note(s):

- 1. Dimensioning & tolerancing conform to ASME Y14.5M-1994.
- 2. All dimensions are in millimeters. Angles are in degrees.
- 3. Dimension b applies to metallized terminal and is measured between 0.25mm and 0.30mm from terminal tip. Dimension L1 represents terminal full back from package edge up to 0.15mm is acceptable.
- 4. Coplanarity applies to the exposed heat slug as well as the terminal.
- 5. Radius on terminal is optional.
- 6. N is the total number of terminals.

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Figure 41: Package Marking

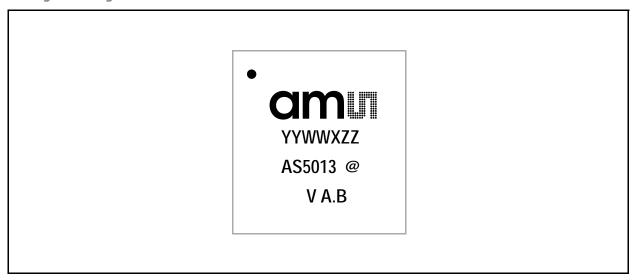


Figure 42:

Package Code: YYWWXZZ@

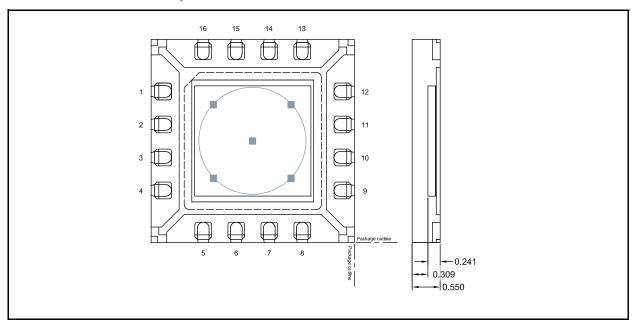
YY	ww	Х	ZZ	@
Last two digits of the current year	Manufacturing Week	Assembly plant identifier	Assembly traceability code	Sublot identifier

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## **Mechanical Data**

Figure 43:
Die Placement and Hall Array Position

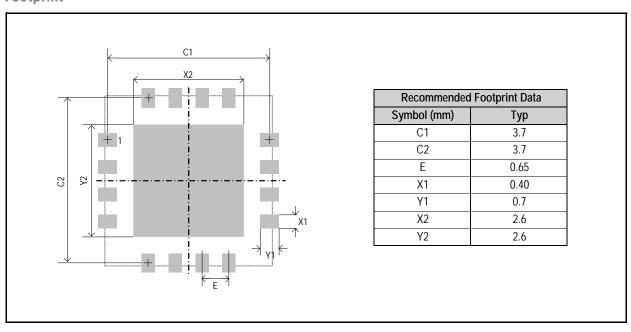


#### Note(s):

- 1. All dimensions in mm.
- 2. The Hall array center is located in the center of the IC package. Hall array radius is 1.1mm.
- 3. Die thickness is 150 $\mu m$  nominal.
- 4. Adhesive thickness 15 $\mu m$  nominal.
- 5. Leadframe thickness 76µm nominal.

## **Recommended Footprint**

Figure 44: Footprint



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# **Recommended Mounting**

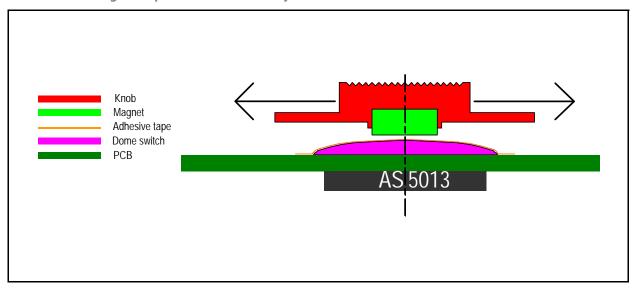
The typical mounting configuration of the AS5013 with the mechanics is on both sides of the PCB:

- Mechanics + Magnet on the top side
- AS5013 IC on the bottom side

A thickness of 0.3mm to 1.0mm for the PCB is recommended.

A dome switch for push button function can be added as well.

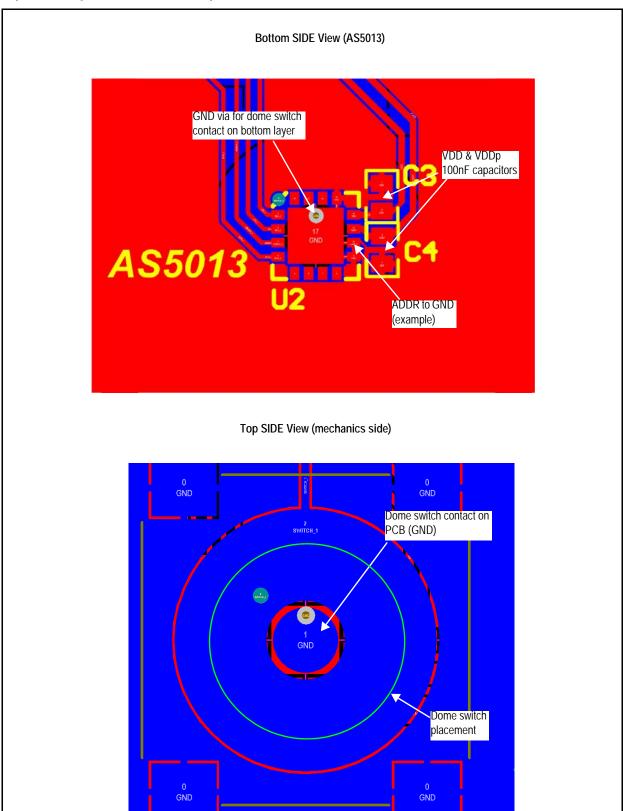
Figure 45: AS5013 Mounting Example for Low Profile Joystick



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Figure 46: Layout Example for Low Profile Joystick



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# Ordering & Contact Information

The devices are available as the standard products shown in Figure 47.

Figure 47: **Ordering Information** 

Ordering Code	Package	Marking	Delivery Form
AS5013-IQFT	16-pin QFN (4mmx4mmx0.55mm)	AS5013	Tape & Reel

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# **Document Status**

Document Status	Product Status	Definition
Product Preview	Pre-Development	Information in this datasheet is based on product ideas in the planning phase of development. All specifications are design goals without any warranty and are subject to change without notice
Preliminary Datasheet	Pre-Production	Information in this datasheet is based on products in the design, validation or qualification phase of development. The performance and parameters shown in this document are preliminary without any warranty and are subject to change without notice
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# **Revision Information**

Changes from 1.15 to current revision 2-00 (2021-Apr-19)	Page
Content was updated to latest <b>ams</b> design	
Added section "Mechanical Data"	42

## Note(s):

- 1. Page and figure numbers for the previous version may differ from page and figure numbers in the current revision.
- 2. Correction of typographical errors is not explicitly mentioned.

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- 5 Absolute Maximum Ratings
- **6 Electrical Characteristics**
- 6 Operating Conditions
- 7 Digital IO pads DC/AC Characteristics

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