

# 10.1 GHz to 11.7 GHz, Low Noise Amplifier

Data Sheet ADL5723

#### **FEATURES**

Frequency range: 10.1 GHz to 11.7 GHz
Typical gain of >24 dB
Low noise input
Noise figure
2.2 dB typical at 10.1 GHz

2.3 dB typical at 11.7 GHz

High linearity input

≥2.2 dBm typical input third-order intercept (IIP3)

-10 dBm typical input 1 dB compression point (P1dB)

Matched 50  $\Omega$  single-ended input
Matched 100  $\Omega$  differential outputs
8-lead, 2.00 mm  $\times$  2.00 mm LFCSP microwave packaging

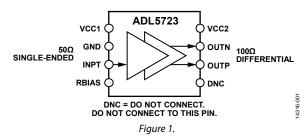
#### **APPLICATIONS**

Point to point microwave radios Instrumentation Satellite communications (SATCOM) Phased arrays

# GENERAL DESCRIPTION

The ADL5723 is a narrow-band, high performance, low noise amplifier (LNA) targeting microwave radio link receiver designs. The monolithic silicon germanium (SiGe) design is optimized for microwave radio link bands ranging from 10.1 GHz to 11.7 GHz. The unique design offers a single-ended 50  $\Omega$  input impedance and provides a 100  $\Omega$  balanced differential output that is ideal for driving Analog Devices, Inc., differential downconverters and radio frequency (RF) sampling analog-to-digital converters

#### **FUNCTIONAL BLOCK DIAGRAM**



(ADCs). This LNA provides noise figure performance that, in the past, required more expensive three-five (III-V) compounds process technology to achieve.

The ADL5721 and ADL5723 to ADL5726 family of narrow-band LNAs are each packaged in a tiny, thermally enhanced, 2.00 mm  $\times$  2.00 mm LFCSP package. The ADL5721 and ADL5723 to ADL5726 family operates over the temperature range of  $-40^{\circ}$ C to  $+85^{\circ}$ C.

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## **REVISION HISTORY**

4/16—Revision 0: Initial Version

# **SPECIFICATIONS**

## **AC SPECIFICATIONS**

 $VCC1 = 1.8 \text{ V}, VCC2 = 3.3 \text{ V}, RBIAS = 442 \ \Omega, T_A = 25 ^{\circ}\text{C}, Z_{SOURCE} = 50 \ \Omega, Z_{LOAD} = 100 \ \Omega \ differential, unless otherwise noted.$ 

Table 1.

Parameter	Test Conditions/Comments	Min	Тур	Max	Unit
FREQUENCY RANGE		10.1		11.7	GHz
FREQUENCY = 10.1 GHz					
Gain (S21)			24.5		dB
Noise Figure			2.2		dB
Input Third-Order Intercept (IIP3)	$\Delta f = 1$ MHz, input power (P <sub>IN</sub> ) = -30 dBm per tone		2.2		dBm
Input 1 dB Compression Point (P1dB)			-10		dBm
Input Return Loss (S11)			10		dB
Output Return Loss (S22)			10		dB
FREQUENCY = 11.7 GHz					
Gain (S21)			24.1		dB
Noise Figure			2.3		dB
Input Third-Order Intercept (IIP3)	$\Delta f = 1$ MHz, $P_{IN} = -30$ dBm per tone		3.5		dBm
Input 1 dB Compression Point (P1dB)			-10		dBm
Input Return Loss (S11)			10		dB
Output Return Loss (S22)			10		dB

## **DC SPECIFICATIONS**

Table 2.

Parameter	Test Conditions/Comments	Min	Тур	Max	Unit
POWER INTERFACE					
Voltage					
VCC1		1.65	1.8	1.95	V
VCC2		3.1	3.3	3.5	V
Quiescent Current vs. Temperature					
VCC1	$T_A = 25$ °C		23.2		mA
	$-40$ °C $\leq$ T <sub>A</sub> $\leq$ $+85$ °C		23.5		mA
VCC2	$T_A = 25$ °C		85.0		mA
	$-40$ °C $\leq$ T <sub>A</sub> $\leq$ $+85$ °C		85.6		mA

## **ABSOLUTE MAXIMUM RATINGS**

Table 3.

Parameter	Rating
Supply Voltages	
VCC1	2.25 V
VCC2	4.1 V
Maximum Junction Temperature	150°C/W
Operating Temperature Range	-40°C to +85°C
Storage Temperature Range	−55°C to +125°C
Lead Temperature Range (Soldering 60 sec)	−65°C to +150°C

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

### THERMAL RESISTANCE

 $\theta_{JA}$  is thermal resistance, junction to ambient (°C/W),  $\theta_{JB}$  is thermal resistance, junction to board (°C/W), and  $\theta_{JC}$  is thermal resistance, junction to case (°C/W).

**Table 4. Thermal Resistance** 

Package Type	$\theta_{JA}^{1}$	$\theta_{JB}^{1}$	$\theta_{JC}^{1}$	Unit
8-Lead LFCSP	39.90	23.88	3.71	°C/W

 $<sup>^{1}</sup>$  See JEDEC standard JESD51-2 for additional information on optimizing the thermal impedance for a printed circuit board (PCB) with 3  $\times$  4 vias.

## **ESD CAUTION**



**ESD** (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

# PIN CONFIGURATION AND FUNCTION DESCRIPTIONS



- NOTES
  1. DNC = DO NOT CONNECT. DO NOT CONNECT TO THIS PIN.
- CONNECT TO THIS PIN.

  2. THE EXPOSED PAD MUST BE SOLDERED TO A LOW IMPEDANCE GROUND PLANE.

  3. THE DEVICE NUMBER ON THE FIGURE DOES NOT INDICATE THE LABEL ON THE PACKAGE. REFER TO THE PIN 1 INDICATOR FOR THE PIN LOCATIONS.

Figure 2. Pin Configuration

**Table 5. Pin Function Descriptions** 

Pin No.	Mnemonic	Description
1	VCC1	1.8 V Power Supply. It is recommended to place the decoupling capacitors as close to this pin as possible.
2	GND	Ground.
3	INPT	RF Input. This is a 50 $\Omega$ single-ended input.
4	RBIAS	Resistor Bias. For typical operation, connect a 442 $\Omega$ resistor from RBIAS to GND. It is recommended to place the RBIAS resistor as close to the pin as possible.
5	DNC	Do Not Connect. Do not connect to this pin.
6, 7	OUTP, OUTN	RF Outputs. These pins are 100 $\Omega$ differential outputs.
8	VCC2	3.3 V Power Supply. It is recommended to place the decoupling capacitors as close to this pin as possible.
	EPAD (EP)	Exposed Pad. The exposed pad must be soldered to a low impedance ground plane.

# TYPICAL PERFORMANCE CHARACTERISTICS

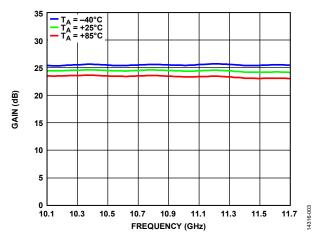


Figure 3. Gain vs. Frequency for Various Temperatures

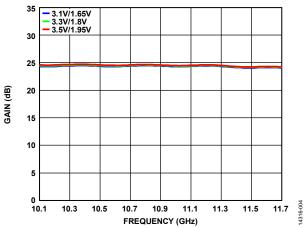


Figure 4. Gain vs. Frequency for Various Supply Voltages

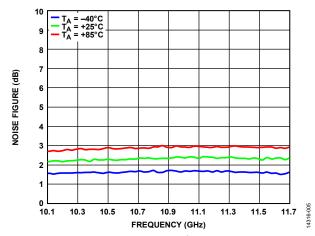


Figure 5. Nosie Figure vs. Frequency for Various Temperatures

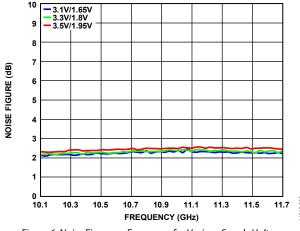


Figure 6. Noise Figure vs. Frequency for Various Supply Voltages

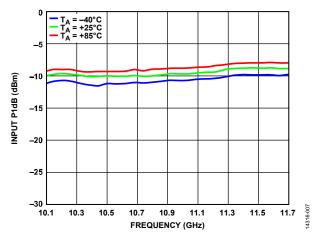


Figure 7. Input P1dB vs. Frequency for Various Temperatures

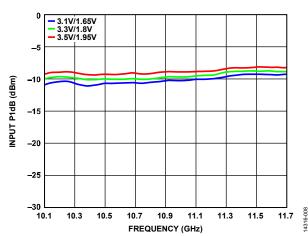


Figure 8. Input P1dB vs. Frequency for Various Supply Voltages

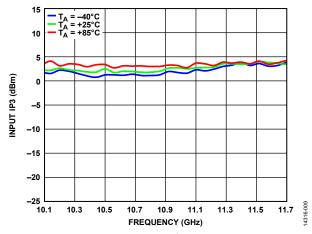


Figure 9. Input IP3 vs. Frequency for Various Temperatures

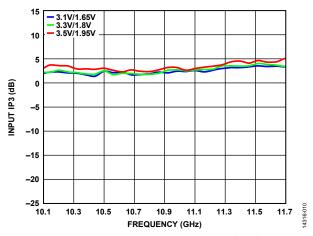


Figure 10. Input IP3 vs. Frequency Various Supply Voltages

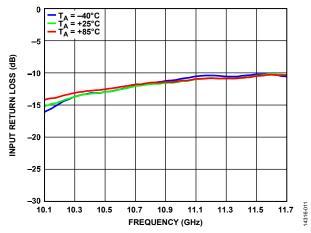


Figure 11. Input Return Loss vs. Frequency for Various Temperatures

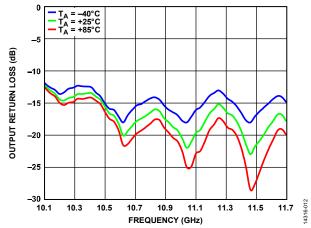


Figure 12. Output Return Loss vs. Frequency for Various Temperatures

## THEORY OF OPERATION

The ADL5723 is a narrow-band, high performance, low noise amplifier targeting microwave radio link receiver designs. The monolithic SiGe design is optimized for microwave radio link bands ranging from 10.1 GHz to 11.7 GHz.

The unique design of the ADL5723 offers a single-ended 50  $\Omega$  input impedance via the INPT pin, and provides a 100  $\Omega$  balanced differential output via the OUTP and OUTN pins.

This LNA is ideal for driving Analog Devices differential downconverters and RF sampling ADCs.

The ADL5723 provides cost-effective noise figure performance without requiring more expensive III-V compounds process technology.

The ADL5723 is available in a 2.00 mm  $\times$  2.00 mm LFCSP package, and operates over the temperature range of  $-40^{\circ}$ C to  $+85^{\circ}$ C.

# APPLICATIONS INFORMATION LAYOUT

Solder the exposed pad on the underside of the ADL5723 to a low thermal and electrical impedance ground plane. This pad is typically soldered to an exposed opening in the solder mask on the evaluation board. Connect the ground vias to all other ground layers on the evaluation board to maximize heat dissipation from the device package.

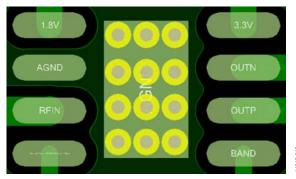


Figure 13. Evaluation Board Layout for the ADL5723 Package

#### **DIFFERENTIAL vs. SINGLE-ENDED OUTPUT**

This section provides the test results that compare the ADL5723 using a differential vs. a single-ended output. When using the device as a single-ended output, use the RFOP output on the evaluation board and terminate RFON to 50  $\Omega$ . Note that the converse can be done as well; however, doing so produces slightly different results from the plots shown in this section because there is some amplitude imbalance between the two differential ports, RFOP and RFON. The output trace and connector loss were not deembedded for these measurements.

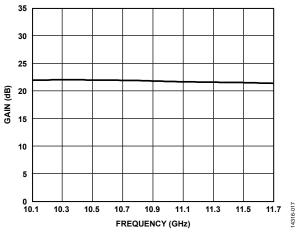


Figure 14. Gain vs. Frequency

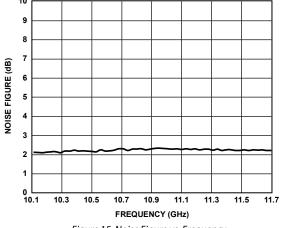


Figure 15. Noise Figure vs. Frequency

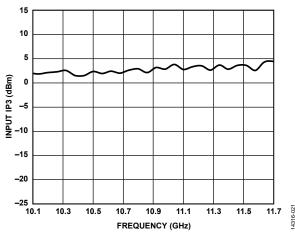


Figure 16. Input IP3 vs. Frequency

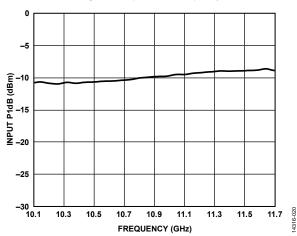


Figure 17. Input P1dB vs. Frequency

## **EVALUATION BOARD**

The ADL5723-EVALZ comes with an ADL5723 chip. It supports a single 5 V supply for ease of use. For 5 V operation, use the 3.3 V and 1.8 V test loops for evaluation purposes only. When using a 3.3 V or 1.8 V supply, remove the R1 and R2 resistors from the evaluation board. Figure 19 shows the ADL5723-EVALZ evaluation board lab bench setup.

### **INITIAL SETUP**

To set up the ADL5723-EVALZ, take the following steps:

- 1. Power up the ADL5723-EVALZ with a 5 V dc supply. The supply current of the evaluation board is approximately 111 mA, which is a combination of the VCC1 (1.8 V) and the VCC2 (3.3 V) currents.
- Connect the signal generator to the input of the ADL5723-EVALZ.
- 3. Connect RFOP and RFON to a 180° hybrid that works within the 10.1 GHz to 11.7 GHz frequency range.
- 4. Connect the difference output of the hybrid to the spectrum analyzer. The sum port of the hybrid must be terminated to 50  $\Omega$ .

See Figure 19 for the ADL5723-EVALZ lab bench setup.

### **RESULTS**

Figure 18 shows the expected results when testing the ADL5723-EVALZ using the Rev. A version of the evaluation board and its software. Note that future iterations of the software may produce different results. See the ADL5723 product page for the most recent software version.

Figure 18 shows the results of the differential output for an input of 10.1~GHz at -15~dBm. The hybrid and board loss have not been deembedded.

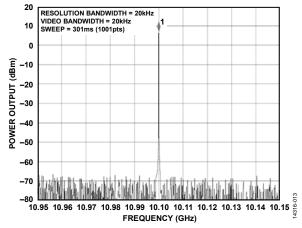


Figure 18. Test Results at 10.1 GHz

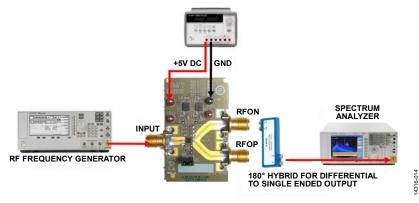


Figure 19. ADL5723-EVALZ Lab Bench Setup

## **BASIC CONNECTIONS FOR OPERATION**

Figure 20 shows the basic connections for operating the ADL5723 as it is implemented on the evaluation board of the device.

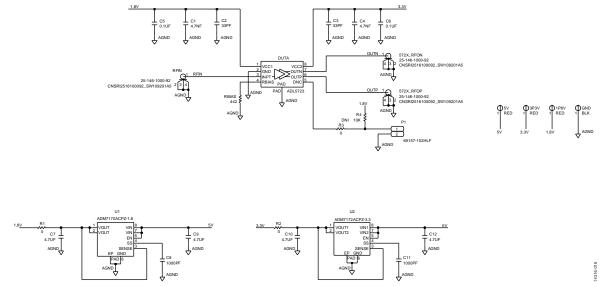


Figure 20. Evaluation Board Schematic

**Table 6. Evaluation Board Configuration Options** 

Component	Function	Default Condition
3P3V, 1P8V, GND, 5V	Power supplies and ground.	Not applicable
RFIN, 572X_RFOP, 572x_RFON	Input, output, and data.	Not applicable
RBIAS	$442 \Omega$ for RBIAS.	RBIAS = $442 \Omega (0402)$
R1, R2	1.8 V and 3.3 V regulator connections.	$R1, R2 = 0 \Omega (0402)$
R3	Do not install (DNI).	R3 = DNI (0402)
R4	Pull-up or pull-down resistor.	$R4 = 10 \text{ k}\Omega (0402)$
C1 to C12	The capacitors provide the required decoupling for the supply related pins.	C1, C4 = 4.7 nF (0402), C2, C3 = 33 pF (0402), C5, C6 = 0.1 µF (0402), C7, C9, C10, C12 = 4.7 µF (0603), C8, C11 = 1000 pF (0603)
P1	Jumper to change bands, 2-pin jumper.	Not applicable
U1	ADM7170ACPZ-1.8 1.8 V regulator.	Not applicable
U2	ADM7172ACPZ-3.3 3.3 V regulator.	Not applicable
DUTA	ADL5723 device under test (DUT).	Not applicable

# **OUTLINE DIMENSIONS**

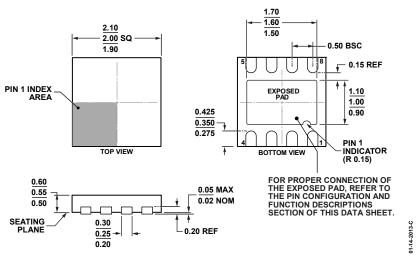


Figure 21. 8-Lead Lead Frame Chip Scale Package [LFCSP] 2.00 mm × 2.00 mm Body, and 0.55 mm Package Height (CP-8-10) Dimensions shown in millimeters

## **ORDERING GUIDE**

Model <sup>1</sup>	Temperature Range	Package Description	Package Option
ADL5723ACPZN-R7	−40°C to +85°C	8-Lead Lead Frame Chip Scale Package [LFCSP]	CP-8-10
ADL5723-EVALZ		Evaluation Board	

<sup>&</sup>lt;sup>1</sup> Z = RoHS-Compliant Part.

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