## Data Sheet

## FEATURES

Single positive supply (self biased)
High OIP2: $\mathbf{5 2 \mathrm { dBm } \text { typical at } 0 . 6 \mathrm { GHz } \text { to } 7 . 5 \mathrm { GHz } , ~ ( 1 )}$
High gain: 15 dB typical at 0.6 GHz to $\mathbf{6 ~ G H z}$
High OIP3: $\mathbf{3 2} \mathbf{d B m}$ typical
Low noise figure: 3.5 dB typical at $\mathbf{0 . 4 ~ \mathbf { ~ G H z } \text { to } \mathbf { 6 ~ G H z }}$
RoHS-compliant, $3 \mathrm{~mm} \times 3 \mathrm{~mm}$, 16-lead LFCSP

## APPLICATIONS

## Test instrumentation <br> Military communications

## FUNCTIONAL BLOCK DIAGRAM


high output second-order intercept (OIP2) of 52 dBm typical at 0.6 GHz to 6 GHz , making the ADL8104 suitable for military and test instrumentation applications.

The ADL8104 also features inputs and outputs that are internally matched to $50 \Omega$. The $\mathrm{RF}_{\text {IN }}$ and $\mathrm{RF}_{\text {out }}$ pins are internally ac-coupled and the bias inductor is also integrated, making the ADL8104 ideal for surface-mounted technology (SMT)-based, high density applications.

The ADL8104 is housed in an RoHS-compliant, $3 \mathrm{~mm} \times 3 \mathrm{~mm}$, 16-lead LFCSP.

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## REVISION HISTORY

9/2020—Revision 0: Initial Version

## SPECIFICATIONS

### 0.4 GHz TO 0.6 GHz FREQUENCY RANGE

$\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}$, total supply current $\left(\mathrm{I}_{\mathrm{DQ}}\right)=150 \mathrm{~mA}, \mathrm{R}_{\text {BIAS }}=90.9 \Omega$, and $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$, unless otherwise noted.
Table 1.

| Parameter | Min | Typ | Max | Unit | Test Conditions/Comments |
| :--- | :--- | :--- | :--- | :--- | :--- |
| FREQUENCY RANGE | 0.4 | 0.6 | GHz |  |  |
| GAIN | 11.5 | 14 | dB |  |  |
| $\quad$ Gain Variation over Temperature |  | 0.036 | $\mathrm{~dB} /{ }^{\circ} \mathrm{C}$ |  |  |
| NOISE FIGURE | 3.5 |  | dB |  |  |
| RETURN LOSS |  |  |  |  |  |
| $\quad$ Input |  | 12 | dB |  |  |
| Output |  | 13 | dB |  |  |
| OUTPUT |  |  |  |  |  |
| OP1dB | 16.5 | 19 | dBm |  |  |
| Saturated Output Power (PSAT) |  | 21 | dBm |  |  |
| OIP3 |  | 32 | dBm | Measurement taken at output power (Pout) per tone $=5 \mathrm{dBm}$ |  |
| OIP2 | 50 | dBm | Measurement taken at Pout per tone $=5 \mathrm{dBm}$ |  |  |
| POWER ADDED EFFICIENCY (PAE) |  | 18 | $\%$ | Measured at $\mathrm{P}_{\text {SAT }}$ |  |

### 0.6 GHz TO 6 GHz FREQUENCY RANGE

$\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}, \mathrm{I}_{\mathrm{DQ}}=150 \mathrm{~mA}, \mathrm{R}_{\text {BIAS }}=90.9 \Omega$, and $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$, unless otherwise noted.
Table 2.

| Parameter | Min | Typ | Max | Unit | Test Conditions/Comments |
| :---: | :---: | :---: | :---: | :---: | :---: |
| FREQUENCY RANGE | 0.6 |  | 6 | GHz |  |
| GAIN <br> Gain Variation over Temperature | 12 | $\begin{aligned} & 15 \\ & 0.030 \end{aligned}$ |  | $\begin{aligned} & \mathrm{dB} \\ & \mathrm{~dB} /{ }^{\circ} \mathrm{C} \end{aligned}$ |  |
| NOISE FIGURE |  | 3.5 |  | dB |  |
| RETURN LOSS Input Output |  | $\begin{aligned} & 12 \\ & 12 \end{aligned}$ |  | $\begin{aligned} & \mathrm{dB} \\ & \mathrm{~dB} \end{aligned}$ |  |
| OUTPUT OP1dB $\mathrm{P}_{\text {sat }}$ OIP3 OIP2 |  | $\begin{aligned} & 20 \\ & 21 \\ & 32 \\ & 52 \end{aligned}$ |  | dBm <br> dBm <br> dBm <br> dBm | Measurement taken at Pout per tone $=5 \mathrm{dBm}$ Measurement taken at Pout per tone $=5 \mathrm{dBm}$ |
| PAE |  | 12 |  | \% | Measured at $\mathrm{P}_{\text {SAT }}$ |

## 6 GHz TO 7.5 GHz FREQUENCY RANGE

$\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}, \mathrm{I}_{\mathrm{DQ}}=150 \mathrm{~mA}, \mathrm{R}_{\text {BIAS }}=90.9 \Omega$, and $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$, unless otherwise noted.
Table 3.

| Parameter | Min | Typ | Max | Unit | Test Conditions/Comments |
| :--- | :--- | :--- | :--- | :--- | :--- |
| FREQUENCY RANGE | 6 |  | 7.5 | GHz |  |
| GAIN | 10 | 13 |  | dB |  |
| $\quad$ Gain Variation over Temperature |  | 0.041 |  | $\mathrm{~dB} /{ }^{\circ} \mathrm{C}$ |  |
| NOISE FIGURE | 4.5 |  | dB |  |  |
| RETURN LOSS |  |  |  |  |  |
| $\quad$ Input | 12 |  | dB |  |  |
| $\quad$ Output |  | 12 | dB |  |  |

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| Parameter | Min | Typ $\quad$ Max | Unit | Test Conditions/Comments |
| :--- | :--- | :--- | :--- | :--- |
| OUTPUT |  |  |  |  |
| OP1dB | 15.5 | 18 |  | dBm |
| PSAT |  | 19 |  | dBm |
| OIP3 | 32 |  | dBm | Measurement taken at Pout per tone $=5 \mathrm{dBm}$ |
| OIP2 | 52 | dBm | Measurement taken at Pout per tone $=5 \mathrm{dBm}$ |  |
| PAE | 12 | $\%$ | Measured at $\mathrm{P}_{\text {SAT }}$ |  |

## DC SPECIFICATIONS

Table 4.

| Parameter | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: |
| SUPPLY CURRENT |  |  |  |  |
| log |  | 150 |  | mA |
| Drain Current (ldo) |  | 144 |  | mA |
| $\mathrm{R}_{\text {bias }}$ Current ( (lrbias) |  | 6 |  | mA |
| SUPPLY VOLTAGE |  |  |  |  |
| $V_{\text {DD }}$ | 3 | 5 | 5.5 | V |

ADL8104

## ABSOLUTE MAXIMUM RATINGS

Table 5.

| Parameter | Rating |
| :---: | :---: |
| $\mathrm{V}_{\mathrm{DD}}$ | 6 V |
| RF Input Power | 25 dBm |
| Continuous Power Dissipation (PDISS), $T_{A}=85^{\circ} \mathrm{C}$ (Derate $22.57 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ Above $85^{\circ} \mathrm{C}$ ) | 2.03 W |
| Temperature |  |
| Storage Range | $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ |
| Operating Range | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ |
| Peak Reflow (Moisture Sensitivity Level 3 (MSL3)) ${ }^{1}$ | $260^{\circ} \mathrm{C}$ |
| Junction to Maintain 1,000,000 Hours Mean Time to Failure (MTTF) | $175^{\circ} \mathrm{C}$ |
| $\begin{aligned} & \text { Nominal Junction }\left(\mathrm{T}_{\mathrm{A}}=85^{\circ} \mathrm{C}, \mathrm{~V}_{\mathrm{DD}}=5 \mathrm{~V}\right. \text {, } \\ & \left.\mathrm{I}_{\mathrm{DQ}}=150 \mathrm{~mA}\right) \end{aligned}$ | $118.22^{\circ} \mathrm{C}$ |

${ }^{1}$ See the Ordering Guide for more information.
Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

## THERMAL RESISTANCE

Thermal performance is directly linked to printed circuit board (PCB) design and operating environment. Close attention to PCB thermal design is required.
$\theta_{\mathrm{JC}}$ is the junction to case thermal resistance.
Table 6. Thermal Resistance

| Package Type | $\boldsymbol{\theta}_{\text {Jc }}$ | Unit |
| :--- | :--- | :--- |
| CP-16-35 | 44.3 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |

## PIN CONFIGURATION AND FUNCTION DESCRIPTIONS



NOTES

1. NO CONNECT. THESE PINS ARE NOT CONNECTED INTERNALLY. THESE PINS MUST BE CONNECTED TO THE RF AND DC GROUND.
2. EXPOSED PAD. THE EXPOSED PAD

MUST BE CONNECTED TO THE RF
AND DC GROUND.
Figure 2. Pin Configuration
Table 8. Pin Function Descriptions

| Pin No. | Mnemonic | Description |
| :--- | :--- | :--- |
| 1,10 | GND | Ground. The GND pin must be connected to the RF and dc ground. See Figure 6 for the interface schematic. |
| 2 | RFIN | RF Input. The RFIN pin is ac-coupled and matched to $50 \Omega$. See Figure 4 for the interface schematic. |
| 3 to 9,12,13, 16 | NC | No Connect. These pins are not connected internally. These pins must be connected to the RF and dc ground. |
| 11 | RFout | RF Output. The RFout pin is ac-coupled and matched to $50 \Omega$. See Figure 5 for the interface schematic. |
| 14 | VDD | Drain Supply Voltage for the Amplifier. See Figure 5 for the interface schematic. |
| 15 | RBAAS | Current Mirror Bias Resistor. Use the RBAS pin to set the quiescent current by connecting an external bias <br> resistor as defined in Table 9 . Refer to Figure 87 for the bias resistor connection. See Figure 3 for the <br> interface schematic. |
|  | EPAD | Exposed Pad. The exposed pad must be connected to the RF and dc ground. |

## INTERFACE SCHEMATICS



Figure 3. R BIAS Interface Schematic

Figure 4. RFis Interface Schematic


Figure 5. $V_{D D}$ and RFout Interface Schematic


Figure 6. GND Interface Schematic

TYPICAL PERFORMANCE CHARACTERISTICS


Figure 7. Gain and Return Loss vs. Frequency, 0.01 GHz to $12 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$, $I_{D Q}=150 \mathrm{~mA}, R_{B I A S}=90.9 \Omega$ (S22 Is the Output Return Loss, S21 Is the Input Return Loss, and S11 Is the Gain)


Figure 8. Gain vs. Frequency for Various Temperatures, 0.3 GHz to 1 GHz, $V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B I A S}=90.9 \Omega$


Figure 9. Gain vs. Frequency for Various $V_{D D}$ and $I_{D Q}$ Values, 0.3 GHz to $1 \mathrm{GHz}, R_{B A A}=90.9 \Omega$


Figure 10. Gain and Return Loss vs. Frequency, 0.1 GHz to $1 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$, $I_{D Q}=150 \mathrm{~mA}, R_{B A A S}=90.9 \Omega$


Figure 11. Gain vs. Frequency for Various Temperatures, 1 GHz to 10 GHz , $V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B I A S}=90.9 \Omega$


Figure 12. Gain vs. Frequency for Various $V_{D D}$ and $I_{D Q}$ Values, 1 GHz to $10 \mathrm{GHz}, R_{\text {BIAS }}=90.9 \Omega$


Figure 13. Gain vs. Frequency for Various $R_{B B A S}$ and $I_{D Q}$ Values, 0.3 GHz to $1 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$


Figure 14. Input Return Loss vs. Frequency for Various Temperatures, 0.3 GHz to $1 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B A A S}=90.9 \Omega$


Figure 15. Input Return Loss vs. Frequency for Various $V_{D D}$ and $I_{D Q}$ Values, 0.3 GHz to $1 \mathrm{GHz}, R_{B A A S}=90.9 \Omega$


Figure 16. Gain vs. Frequency for Various $R_{B B A S}$ and $I_{D Q}$ Values, 1 GHz to $10 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$


Figure 17. Input Return Loss vs. Frequency for Various Temperatures, 1 GHz to $10 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B A A}=90.9 \Omega$


Figure 18. Input Return Loss vs. Frequency for Various $V_{D D}$ and $I_{D Q}$ Values, 1 GHz to $10 \mathrm{GHz}, R_{\text {BIAS }}=90.9 \Omega$


Figure 19. Input Return Loss vs. Frequency for Various $R_{B I A S}$ and $I_{D Q}$ Values, 0.3 GHz to $1 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$


Figure 20. Output Return Loss vs. Frequency for Various Temperatures, 0.3 GHz to $1 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B A A}=90.9 \Omega$


Figure 21. Output Return Loss vs. Frequency for Various VDD and I $I_{D Q}$ Values, 0.3 GHz to $1 \mathrm{GHz}, R_{\text {BIAS }}=90.9 \Omega$


Figure 22. Input Return Loss vs. Frequency for Various $R_{B I A S}$ and $I_{D Q}$ Values, 1 GHz to $10 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$


Figure 23. Output Return Loss vs. Frequency for Various Temperatures, 1 GHz to $10 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B I A S}=90.9 \Omega$


Figure 24. Output Return Loss vs. Frequency for Various $V_{D D}$ and $I_{D Q}$ Values, 1 GHz to $10 \mathrm{GHz}, R_{B A A S}=90.9 \Omega$


Figure 25. Output Return Loss vs. Frequency for Various $R_{B I A S}$ and $I_{D Q}$ Values, 0.3 GHz to $1 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$


Figure 26. Reverse Isolation vs. Frequency for Various Temperatures, 0.3 GHz to $1 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B I A S}=90.9 \Omega$


Figure 27. Reverse Isolation vs. Frequency for Various $V_{D D}$ and $I_{D Q}$ Values, 0.3 GHz to $1 \mathrm{GHz}, R_{B A A}=90.9 \Omega$


Figure 28. Output Return Loss vs. Frequency for Various $R_{B I A S}$ and $I_{D Q}$ Values, 1 GHz to $10 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$


Figure 29. Reverse Isolation vs. Frequency for Various Temperatures, 1 GHz to $10 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B I A S}=90.9 \Omega$


Figure 30. Reverse Isolation vs. Frequency for Various $V_{D D}$ and $I_{D Q}$ Values, 1 GHz to $10 \mathrm{GHz}, R_{B I A S}=90.9 \Omega$


Figure 31. Reverse Isolation vs. Frequency for Various $R_{B I A S}$ and $I_{D Q}$ Values, 0.3 GHz to $1 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$


Figure 32. Noise Figure vs. Frequency for Various Temperatures, 0.3 GHz to $1 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B A A}=90.9 \Omega$


Figure 33. Noise Figure vs. Frequency for Various $V_{D D}$ and $I_{D Q}$ Values, 0.3 GHz to $1 \mathrm{GHz}, R_{B I A S}=90.9 \Omega$


Figure 34. Reverse Isolation vs. Frequency for Various $R_{B I A S}$ and $I_{D Q}$ Values, 1 GHz to $10 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$


Figure 35. Noise Figure vs. Frequency for Various Temperatures, 1 GHz to $10 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B A A S}=90.9 \Omega$


Figure 36. Noise Figure vs. Frequency for Various $V_{D D}$ and $I_{D Q}$ Values, 1 GHz to $10 \mathrm{GHz}, R_{\text {BIAS }}=90.9 \Omega$


Figure 37. Noise Figure vs. Frequency for Various $R_{B I A S}$ and $I_{D Q}$ Values, 0.3 GHz to $1 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$


Figure 38. OP1dB vs. Frequency for Various Temperatures, 0.35 GHz to $10 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B I A S}=90.9 \Omega$


Figure 39. OP1dB vs. Frequency for Various Temperatures, 0.35 GHz to $1 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B A A}=90.9 \Omega$


Figure 40. Noise Figure vs. Frequency for Various $R_{B I A S}$ and $I_{D Q}$ Values, 1 GHz to $10 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$


Figure 41. PsAT Vs. Frequency for Various Temperatures, 0.35 GHz to 10 GHz , $V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B A A S}=90.9 \Omega$


Figure 42. OP1dB vs. Frequency for Various Temperatures, 1 GHz to 10 GHz , $V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B A A S}=90.9 \Omega$


Figure 43. $O P 1 d B$ vs. Frequency for Various $V_{D D}$ and $I_{D Q}$ Values, 0.35 GHz to $1 \mathrm{GHz}, R_{\text {BIAS }}=90.9 \Omega$


Figure 44. OP1dB vs. Frequency for Various $R_{B I A S}$ and $I_{D Q}$ Values, 0.35 GHz to $1 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$


Figure 45. PSAT vs. Frequency for Various Temperatures, 0.35 GHz to 1 GHz , $V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B I A S}=90.9 \Omega$


Figure 46. $O P 1 d B$ vs. Frequency for Various $V_{D D}$ and $I_{D Q}$ Values, 1 GHz to $10 \mathrm{GHz}, R_{\text {BIAS }}=90.9 \Omega$


Figure 47. OP1dB vs. Frequency for Various RBBAs and IDO Values, 1 GHz to $10 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$


Figure 48. $P_{S A T}$ Vs. Frequency for Various Temperatures, 1 GHz to 10 GHz , $V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B I A S}=90.9 \Omega$


Figure 49. $P_{S A T}$ Vs. Frequency for Various $V_{D D}$ and $I_{D Q}$ Values, 0.35 GHz to $1 \mathrm{GHz}, R_{\text {BIAS }}=90.9 \Omega$


Figure 50. PSAT Vs. Frequency for Various RBAAs and IDQ Values, 0.35 GHz to $1 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$


Figure 51. PAE vs. Frequency for Various Temperatures, 0.35 GHz to 1 GHz , $V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B A A}=90 \Omega$


Figure 52. $P_{S A T}$ Vs. Frequency for Various $V_{D D}$ and $I_{D Q}$ Values, 1 GHz to 10 GHz , $R_{B A A}=90.9 \Omega$


Figure 53. PSAT Vs. Frequency for Various $R_{B A A S}$ and $I_{D Q}$ Values, 1 GHz to $10 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$


Figure 54. PAE vs. Frequency for Various Temperatures, 1 GHz to 10 GHz , $V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B A A}=90 \Omega$


Figure 55. PAE vs. Frequency for Various $V_{D D}$ and $I_{D Q}$ Values, 0.35 GHz to $1 \mathrm{GHz}, R_{\text {BIAS }}=90.9 \Omega$


Figure 56. PAE vs. Frequency for Various RBAAs and IDQ Values, 0.3 GHz to $1 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$


Figure 57. Pout, Gain, PAE, and $I_{D D}$ vs. Input Power, Power Compression at $0.4 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}, R_{B I A S}=90.9 \Omega$


Figure 58. PAE vs. Frequency for Various VDD and IDQ Values, 1 GHz to 10 GHz , $R_{B A A S}=90.9 \Omega$


Figure 59. PAE vs. Frequency for Various $R_{B B A S}$ and $I_{D Q}$ Values, 1 GHz to $10 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$


Figure 60. Pout, Gain, PAE, and IDD vs. Input Power,
Power Compression at $2 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}, R_{B A A S}=90.9 \Omega$


Figure 61. Pout, Gain, PAE, and I IDD vs. Input Power, Power Compression at $5 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}, R_{B I A S}=90.9 \Omega$


Figure 62. OP1dB, Gain, $P_{S A T}$, and $I_{D D}$ (Measured at $P_{S A T}$ ) vs. Supply Voltage, Power Compression at $0.4 \mathrm{GHz}, R_{B I A S}=90.9 \Omega$


Figure 63. OP1dB, Gain, $P_{S A T,}$ and $I_{D D}$ (Measured at $P_{S A T}$ ) vs. Supply Voltage, Power Compression at $5 \mathrm{GHz}, \mathrm{R}_{\mathrm{BIAS}}=90.9 \Omega$


Figure 64. Pout, Gain, PAE, and $I_{D D}$ vs. Input Power, Power Compression at $7 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}, R_{B I A S}=90.9 \Omega$


Figure 65. OP1dB, Gain, $P_{S A T}$, and $I_{D D}$ (Measured at $P_{S A T}$ ) vs. Supply Voltage, Power Compression at $2 \mathrm{GHz}, R_{\text {BIAS }}=90.9 \Omega$


Figure 66. OP1dB, Gain, $P_{S A T}$, and IDD (Measured at $P_{S A T}$ ) vs. Supply Voltage, Power Compression at $7 \mathrm{GHz}, R_{B I A S}=90.9 \Omega$


Figure 67. PDISS Vs. Input Power at $T_{A}=85^{\circ} \mathrm{C}, V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}$, $R_{\text {BIAS }}=90.9 \Omega$


Figure 68. OIP3 vs. Frequency for Various Temperatures, 0.35 GHz to 1 GHz , $V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B I A S}=90.9 \Omega$, Pout per Tone $=5 \mathrm{dBm}$


Figure 69. OIP3 vs. Frequency for Various $V_{D D}$ and $I_{D Q}$ Values, 0.3 GHz to $1 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$, Pout per Tone $=5 \mathrm{dBm}$


Figure 70. OIP3 vs. Frequency for Various Pout per Tone, $V_{D D}=5 \mathrm{~V}, R_{B A S}=90.9 \Omega$, $I_{D Q}=150 \mathrm{~mA}$


Figure 71. OIP3 vs. Frequency for Various Temperatures, 1 GHz to 10 GHz, $V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B I A S}=90.9 \Omega$, POUT Per Tone $=5 \mathrm{dBm}$


Figure 72. OIP3 vs. Frequency for Various $V_{D D}$ and $I_{D Q}$ Values, 1 GHz to $10 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$, Pout per Tone $=5 \mathrm{dBm}$


Figure 73. OIP3 vs. Frequency for Various $R_{B I A S}$ and $I_{D Q}$ Values, 0.3 GHz to $1 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$, Pout per Tone $=5 \mathrm{dBm}$


Figure 74. Third-Order Intermodulation Distortion Relative to Carrier (IMD3) vs. Frequency for Various Pout per Tone, $V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B A S}=90.9 \Omega$


Figure 75. OIP2 vs. Frequency for Various Temperatures, 0.35 GHz to 1 GHz , $V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B I A S}=90.9 \Omega$, Pout per Tone $=5 \mathrm{dBm}$


Figure 76. OIP3 vs. Frequency for Various $R_{B A A}$ and $I_{D Q}$ Values, 1 GHz to $10 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$, Pout per Tone $=5 \mathrm{dBm}$


Figure 77. OIP2 vs. Frequency for Various Pout per Tone, VDD $=5 \mathrm{~V}$, $I_{D Q}=150 \mathrm{~mA}, R_{B I A S}=90.9 \Omega$


Figure 78. OIP2 vs. Frequency for Various Temperatures, 1 GHz to 10 GHz , $V_{D D}=5 \mathrm{~V}, I_{D Q}=150 \mathrm{~mA}, R_{B A A S}=90.9 \Omega$, P OUT per Tone $=5 \mathrm{dBm}$


Figure 79. OIP2 vs. Frequency for Various $V_{D D}$ and $I_{D Q}$ Values, 0.35 GHz to $1 \mathrm{GHz}, R_{\text {BIAS }}=90.9 \Omega$, Pout per Tone $=5 \mathrm{dBm}$


Figure 80. OIP2 vs. Frequency for Various RBIAS and IDQ Values, 0.3 GHz to $1 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$, Pout per Tone $=5 \mathrm{dBm}$


Figure 81. IDQ Vs. Input Power for Various Frequencies, $V_{D D}=5 \mathrm{~V}, R_{B A A}=90.9 \Omega$


Figure 82. OIP2 vs. Frequency for Various VDD and $I_{D O}$ Values, 1 GHz to $10 \mathrm{GHz}, R_{\text {BIAS }}=90.9 \Omega$, POUT per Tone $=5 \mathrm{dBm}$


Figure 83. OIP2 vs. Frequency for Various $R_{B A A}$ and $I_{D Q}$ Values, 1 GHz to $10 \mathrm{GHz}, V_{D D}=5 \mathrm{~V}$, Pоит per Tone $=5 \mathrm{dBm}$


Figure 84. $I_{D Q}$ vs. Supply Voltage, $R_{B B A S}=90.9 \Omega$


Figure 85. IDQ vs. Bias Resistor Value, VDD $=5 \mathrm{~V}$

## THEORY OF OPERATION

The ADL8104 is a GaAs, MMIC, pHEMT, low noise wideband amplifier with integrated ac-coupling capacitors and a bias inductor. Figure 86 shows a simplified schematic.
The ADL8104 has ac-coupled, single-ended input and output ports with impedances that are nominally equal to $50 \Omega$ over the 0.4 GHz to 7.5 GHz frequency range. No external matching
components are required. To adjust the quiescent current, connect an external resistor between the RBIAS and VDD pins.


## APPLICATIONS INFORMATION

The basic connections for operating the ADL8104 over the specified frequency range are shown in Figure 87. No external biasing inductor is required, allowing the 5 V supply to be connected to the $\mathrm{V}_{\mathrm{DD}}$ pin. The $1 \mu \mathrm{~F}$ and 1000 pF power supply decoupling capacitors are recommended. The power supply decoupling capacitors shown in Figure 87 represent the configuration used to characterize and qualify the ADL8104.
To set $\mathrm{I}_{\mathrm{DQ}}$, connect a resistor, R 1 , between the $\mathrm{R}_{\text {BIA }}$ and $\mathrm{V}_{\mathrm{DD}}$ pins. A default value of $90.9 \Omega$ is recommended, which results in a nominal $I_{D Q}$ of 150 mA . Table 9 shows how the $I_{D Q}$ and $I_{D D}$ varies vs. the bias resistor value. The $\mathrm{R}_{\text {bias }}$ pin also draws a current that varies with the value of $\mathrm{R}_{\text {BIAS }}$ (see Table 9). Do not leave the $R_{\text {bias }}$ pin open.


## RECOMMENDED BIAS SEQUENCING

See the ADL8104-EVALZ user guide for the recommended bias sequencing information.

Table 9. Recommended Bias Resistor Values

| RBAAS $\boldsymbol{( \Omega )}$ | IDQ $(\mathbf{m A})$ | IDD $(\mathbf{m A})$ | IRBIAS $^{(m A)}$ |
| :--- | :--- | :--- | :--- |
| 0 | 165 | 157.3 | 7.7 |
| 90 | 150 | 144 | 6 |
| 440 | 125 | 120 | 5 |
| 1180 | 100 | 97 | 3 |

## OUTLINE DIMENSIONS



COMPLIANT TO JEDEC STANDARDS MO-220-WEED-2


Figure 88. 16-Lead Lead Frame Chip Scale Package [LFCSP]
$3 \mathrm{~mm} \times 3 \mathrm{~mm}$ Body and 0.75 mm Package Height (CP-16-35)
Dimensions shown in millimeters

ORDERING GUIDE

| Model $^{1,2}$ | Temperature Range $^{\text {² }}$ | MSL Rating |  |  |
| :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |
| ADL8104ACPZN | Package Description $^{4}$ | Package Option |  |  |
| ADL8104ACPZN-R7 | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | MSL3 | 16 -Lead Lead Frame Chip Scale Package [LFCSP] |
| ADL8104-EVALZ |  |  | 16-Lead Lead Frame Chip Scale Package [LFCSP] | CP-16-35 |

${ }^{1}$ The ADL8104ACPZN, ADL8104ACPZN-R7, and ADL8104-EVALZ are RoHS compliant parts.
${ }^{2}$ When ordering the evaluation board only, reference the model number, ADL8104-EVALZ.
${ }^{3}$ See the Absolute Maximum Ratings section for additional information.
${ }^{4}$ The lead finish of the ADL8104ACPZN and ADL8104ACPZN-R7 is nickel palladium gold (NiPdAu).

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