## Low Cost, Voltage Output, High-Side, Current-Sense Amplifier

## FEATURES

Low cost, compact, current-sense solution
3 available gain versions
20 V/V (ADM4073T)
50 V/V (ADM4073F)
100 V/V (ADM4073H)
Typical $\pm 1.0 \%$ full-scale accuracy
Supply current: $500 \boldsymbol{\mu} \mathrm{~A}$
Wide bandwidth: 1.8 MHz
Operating supply: 3 V to 28 V
Wide common-mode range: 2 V to 28 V
Independent of supply voltage
Operating temperature range: $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$
Available in a 6-lead SOT-23 package
Pin-to-pin compatibility with the MAX4073

## APPLICATIONS

## Cell phones

PDAs
Notebook computers
Portable, battery-powered systems
Smart battery packs and chargers
Automotive
Power management systems
PA bias control
General system-level, board-level current monitoring
Precision current sources

## GENERAL DESCRIPTION

The ADM4073 is a low cost, high-side, current-sense amplifier ideal for small portable applications, such as cell phones, notebook computers, PDAs, and other systems where current monitoring is required. The device is available in three different gain models, eliminating the need for gain-setting resistors. Because the ground path is not interrupted, the ADM4073 is particularly useful in rechargeable battery-powered systems, while its wide 1.8 MHz bandwidth makes it suitable for use inside battery-charger control loops. The input common-mode range of 2 V to 28 V is independent of the supply voltage.

FUNCTIONAL BLOCK DIAGRAM


Figure 1.

APPLICATION DIAGRAM


Figure 2.

The voltage on the output pin is determined by the current flowing through the selectable external sense resistor and the gain of the version selected. The operating range is 3 V to 28 V with a typical supply current of $500 \mu \mathrm{~A}$.

The ADM4073 is available in a 6-lead SOT-23 package and is specified over the automotive operating temperature range $\left(-40^{\circ} \mathrm{C}\right.$ to $\left.+125^{\circ} \mathrm{C}\right)$.

Rev. A
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## ADM4073

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10/08-Rev. 0 to Rev. A
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## SPECIFICATIONS

$\mathrm{V}_{\mathrm{RS}+}=2 \mathrm{~V}$ to $28 \mathrm{~V}, \mathrm{~V}_{\text {SENSE }}=\left(\mathrm{V}_{\mathrm{RS}+}-\mathrm{V}_{\mathrm{RS}-}\right)=0 \mathrm{~V}, \mathrm{~V}_{\mathrm{CC}}=3 \mathrm{~V}$ to $28 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$, unless otherwise noted. Typical values are at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C} .{ }^{1}$
Table 1.

| Parameter | Min | Typ | Max | Unit | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: |
| POWER SUPPLY |  |  |  |  |  |
| Operating Voltage Range, V Vcc | 3 |  | 28 | V | Inferred from PSRR test |
| Common-Mode Input Range, $\mathrm{V}_{\text {cmi }}$ | 2 |  | 28 | V | Inferred OUT voltage error test |
| Common-Mode Input Rejection, CMR |  | 90 |  | dB | $\mathrm{V}_{\text {SENSE }}=100 \mathrm{mV}, \mathrm{V}_{\text {cc }}=12 \mathrm{~V}$ |
| Supply Current, Icc |  | 0.5 | 1.2 | mA | $\mathrm{V}_{\text {cc }}=28 \mathrm{~V}$ |
| Leakage Current, IRS+/IRS- |  | 0.05 | 2 | $\mu \mathrm{A}$ | $\mathrm{V}_{\mathrm{cC}}=0 \mathrm{~V}, \mathrm{~V}_{\mathrm{RS}+}=28 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=85^{\circ} \mathrm{C}$ |
| Input Bias Current, $\mathrm{IRS}^{+}$ |  | 20 | 60 | $\mu \mathrm{A}$ |  |
| Input Bias Current, IRS- |  | 40 | 120 | $\mu \mathrm{A}$ |  |
| Full-Scale Sense Voltage, V ${ }_{\text {sense }}$ |  | 150 |  | mV | $\mathrm{V}_{\text {SENSE }}=\left(\mathrm{V}_{\text {RS }+}-\mathrm{V}_{\text {RS }-}\right)$ |
| Total Output Voltage Error ${ }^{2}$ |  |  |  | \% | $\mathrm{V}_{\text {SENSE }}=100 \mathrm{mV}, \mathrm{V}_{\text {cC }}=12 \mathrm{~V}, \mathrm{~V}_{\text {RS }+}=2 \mathrm{~V}$ |
|  |  | $\pm 1.0$ | $\pm 5.0$ | \% | $\mathrm{V}_{\text {SENSE }}=100 \mathrm{mV}, \mathrm{V}_{\text {cc }}=12 \mathrm{~V}, \mathrm{~V}_{\text {RS }+}=12 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ |
|  |  |  | $\pm 5.0$ | \% | $\mathrm{V}_{\text {SENSE }}=100 \mathrm{mV}, \mathrm{V}_{\mathrm{CC}}=12 \mathrm{~V}, \mathrm{~V}_{\text {RS }}=12 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ |
|  |  | $\pm 1.0$ | $\pm 5.0$ | \% | $V_{\text {SENSE }}=100 \mathrm{mV}, \mathrm{V}_{\text {cc }}=28 \mathrm{~V}, \mathrm{~V}_{\text {RS }}=28 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ |
|  |  |  | $\pm 5.0$ | \% | $\mathrm{V}_{\text {SENSE }}=100 \mathrm{mV}, \mathrm{V}_{\text {cC }}=28 \mathrm{~V}, \mathrm{~V}_{\text {RS }}=28 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ |
|  |  | $\pm 7.5$ |  | \% | $\mathrm{V}_{\text {SENSE }}=6.25 \mathrm{mV},^{3} \mathrm{~V}_{\text {CC }}=12 \mathrm{~V}, \mathrm{~V}_{\text {RS }}=12 \mathrm{~V}$ |
| Extrapolated Input Offset Voltage, Vos |  | 1.0 |  | mV | $\mathrm{V}_{\text {CC }}=\mathrm{V}_{\text {RS }}=12 \mathrm{~V}$, $\mathrm{V}_{\text {SENSE }}>10 \mathrm{mV}$ |
| Output High Voltage ( $\mathrm{V}_{\text {cc }}-\mathrm{V}_{\text {OH }}$ ) |  | 0.8 | 1.2 | V | $\mathrm{V}_{\text {CC }}=3 \mathrm{~V}, \mathrm{~V}_{\text {SENSE }}=150 \mathrm{mV}$ (ADM4073T) |
|  |  | 0.8 | 1.2 | V | $\mathrm{V}_{\text {cc }}=7.5 \mathrm{~V}, \mathrm{~V}_{\text {SENSE }}=150 \mathrm{mV}$ (ADM4073F) |
|  |  | 0.8 | 1.2 | V | $\mathrm{V}_{\text {CC }}=15 \mathrm{~V}, \mathrm{~V}_{\text {SENSE }}=150 \mathrm{mV}(\mathrm{ADM} 4073 \mathrm{H}), \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ |
| DYNAMIC CHARACTERISTICS |  |  |  |  |  |
| Bandwidth, BW |  | 1.8 |  | MHz | $\mathrm{V}_{\text {SENSE }}=100 \mathrm{mV}, \mathrm{V}_{\text {cc }}=12 \mathrm{~V}, \mathrm{~V}_{\text {RS }}=12 \mathrm{~V}, \mathrm{C}_{\text {LoAd }}=5 \mathrm{pF}$ ( $\mathrm{ADM4073T}$ ) |
|  |  | 1.7 |  | MHz | $\mathrm{V}_{\text {SENSE }}=100 \mathrm{mV}, \mathrm{V}_{\text {CC }}=12 \mathrm{~V}, \mathrm{~V}_{\text {RS }}=12 \mathrm{~V}, \mathrm{C}_{\text {LOAD }}=5 \mathrm{pF}$ ( $\mathrm{ADM4073F}$ ) |
|  |  | 1.6 |  | MHz | $\mathrm{V}_{\text {SENSE }}=100 \mathrm{mV}, \mathrm{V}_{\text {CC }}=12 \mathrm{~V}, \mathrm{~V}_{\text {RS }+}=12 \mathrm{~V}, \mathrm{C}_{\text {LOAD }}=5 \mathrm{pF}(\mathrm{ADM4073H})$ |
|  |  | 600 |  | kHz | $\mathrm{V}_{\text {SENSE }}=6.25 \mathrm{mV},{ }^{3} \mathrm{~V}_{\text {cc }}=12 \mathrm{~V}, \mathrm{~V}_{\text {RS }}=12 \mathrm{~V}, \mathrm{C}_{\text {LOAD }}=5 \mathrm{pF}(\mathrm{ADM4073T} / \mathrm{F} / \mathrm{H})$ |
| Gain, $\mathrm{A}_{v}$ |  | 20 |  | V/V | ADM4073T |
|  |  | 50 |  | V/V | ADM4073F |
|  |  | 100 |  | V/V | ADM4073H |
| Gain Accuracy |  | $\pm 1.0$ | $\pm 2.0$ | \% | $\begin{aligned} & \mathrm{V}_{\text {SENSE }}=10 \mathrm{mV} \text { to } 150 \mathrm{mV}, \mathrm{~V}_{\mathrm{CC}}=12 \mathrm{~V}, \mathrm{~V}_{\mathrm{RS}+}=12 \mathrm{~V}, \\ & \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}(\mathrm{ADM} 4073 \mathrm{~T} / \mathrm{F}) \end{aligned}$ |
|  |  |  | $\pm 2.0$ | \% | $V_{\text {SENSE }}=10 \mathrm{mV}$ to $150 \mathrm{mV}, \mathrm{V}_{\mathrm{CC}}=12 \mathrm{~V}, \mathrm{~V}_{\text {RS }+}=12 \mathrm{~V}$, $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ (ADM4073T/F) |
|  |  | $\pm 1.0$ | $\pm 1.5$ | \% | $\begin{aligned} & \mathrm{V}_{\text {SENSE }}=10 \mathrm{mV} \text { to } 100 \mathrm{mV}, \mathrm{~V}_{\mathrm{CC}}=12 \mathrm{~V}, \mathrm{~V}_{\mathrm{RS}+}=12 \mathrm{~V}, \\ & \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}(\mathrm{ADM} 4073 \mathrm{H}) \end{aligned}$ |
|  |  |  | $\pm 3.0$ | \% | $\begin{aligned} & \mathrm{V}_{\text {SENSE }}=10 \mathrm{mV} \text { to } 100 \mathrm{mV}, \mathrm{~V}_{\mathrm{CC}}=12 \mathrm{~V}, \mathrm{~V}_{\mathrm{RS}+}=12 \mathrm{~V}, \\ & \mathrm{~T}_{\mathrm{A}}=-40^{\circ} \mathrm{C} \text { to }+125^{\circ} \mathrm{C}(\mathrm{ADM} 4073 \mathrm{H}) \end{aligned}$ |
| OUT Settling Time to $1 \%$ of Final Value |  | 400 |  | ns | $\mathrm{V}_{\text {SENSE }}=6.25 \mathrm{mV}$ to $100 \mathrm{mV}, \mathrm{V}_{\mathrm{CC}}=12 \mathrm{~V}, \mathrm{~V}_{\mathrm{RS}+}=12 \mathrm{~V}, \mathrm{C}_{\text {LOAD }}=5 \mathrm{pF}$ |
|  |  | 800 |  | ns | $\mathrm{V}_{\text {SENSE }}=100 \mathrm{mV}$ to $6.25 \mathrm{mV}, \mathrm{V}_{\text {CC }}=12 \mathrm{~V}, \mathrm{~V}_{\text {RS }+}=12 \mathrm{~V}, \mathrm{C}_{\text {LOAD }}=5 \mathrm{pF}$ |
| Output Resistance, Rout |  | 12 |  | $\mathrm{k} \Omega$ |  |
| Power Supply Rejection Ratio, PSRR |  | 78 |  | dB | $\mathrm{V}_{\text {SENSE }}=60 \mathrm{mV}, \mathrm{V}_{\text {cc }}=3 \mathrm{~V}$ to 28V (ADM4073T) |
|  |  | 85 |  | dB | $\mathrm{V}_{\text {SENSE }}=24 \mathrm{mV}, \mathrm{V}_{\text {cC }}=3 \mathrm{~V}$ to 28V (ADM4073F) |
|  |  | 90 |  | dB | $\mathrm{V}_{\text {SENSE }}=12 \mathrm{mV}, \mathrm{V}_{\text {CC }}=3 \mathrm{~V}$ to 28V (ADM4073H) |
| Power-Up Time ${ }^{4}$ |  | 5 |  | $\mu \mathrm{s}$ | $\mathrm{C}_{\text {LOAd }}=5 \mathrm{pF}, \mathrm{V}_{\text {SENSE }}=100 \mathrm{mV}$ |
| Saturation Recovery Time ${ }^{5}$ |  | 5 |  | $\mu \mathrm{s}$ | $\mathrm{C}_{\text {LOAD }}=5 \mathrm{pF}, \mathrm{V}_{\text {CC }}=12 \mathrm{~V}, \mathrm{~V}_{\text {RS }+}=12 \mathrm{~V}$ |

[^0]
## ABSOLUTE MAXIMUM RATINGS

Table 2.

| Parameter | Rating |
| :--- | :--- |
| $\mathrm{V}_{\mathrm{CC}}$ to GND | -0.3 V to +30 V |
| RS,+ RS- to GND | -0.3 V to +30 V |
| OUT to GND | -0.3 V to $\left(\mathrm{V}_{\mathrm{CC}}+0.3 \mathrm{~V}\right)$ |
| OUT Short-Circuit to GND | Continuous |
| Differential Input Voltage (V $\left.\mathrm{V}_{\text {R }+}-\mathrm{V}_{\mathrm{RS}-}\right)$ | $\pm 5 \mathrm{~V}$ |
| Current into Any Pin | $\pm 20 \mathrm{~mA}$ |
| Storage Temperature Range | $-65^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ |
| Operating Temperature Range | $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ |
| Lead Temperature, Soldering (10 sec) | $300^{\circ} \mathrm{C}$ |
| Junction Temperature | $150^{\circ} \mathrm{C}$ |

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## THERMAL CHARACTERISTICS

$\theta_{\mathrm{JA}}$ is specified for the worst-case conditions, that is, a device soldered in a circuit board for surface-mount packages.

Table 3. Thermal Resistance

| Package Type | $\boldsymbol{\theta}_{\mathrm{JA}}$ | Unit |
| :--- | :--- | :--- |
| 6-Lead SOT-23 | 169.5 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |

## ESD CAUTION

|  | ESD (electrostatic discharge) sensitive device. <br> Charged devices and circuit boards can discharge <br> without detection. Although this product features <br> patented or proprietary protection circuitry, damage <br> may occur on devices subjected to high energy ESD. <br> Therefore, proper ESD precautions should be taken to <br> avoid performance degradation or loss of functionality. |
| :--- | :--- |

## PIN CONFIGURATION AND FUNCTION DESCRIPTIONS



Table 4. Pin Function Descriptions

| Pin No. | Mnemonic | Description |
| :--- | :--- | :--- |
| 1 | GND | Chip Ground Pin. |
| 2 | GND | Chip Ground Pin. |
| 3 | Vcc | Chip Power Supply. Requires a 0.1 $\mu$ F capacitor to ground. |
| 4 | RS + | Power-Side Connection to the External Sense Resistor. |
| 5 | RS- | Load-Side Connection to the External Sense Resistor. |
| 6 | OUT | Voltage Output. Vout is proportional to V Vense. Output impedance is approximately $12 \mathrm{k} \Omega$. |

## ADM4073

## TYPICAL PERFORMANCE CHARACTERISTICS



Figure 4. Supply Current vs. Supply Voltage ( $V_{\text {SENSE }}=6.25 \mathrm{mV}$ )


Figure 5. Supply Current vs. Supply Voltage ( $V_{\text {SENSE }}=100 \mathrm{mV}$ )


Figure 6. Supply Current vs. Temperature


Figure 7. Supply Current vs. $R S+$ Voltage $\left(V_{\text {SENSE }}=6.25 \mathrm{mV}\right)$


Figure 8. Supply Current vs. RS+Voltage (VSENSE $=100 \mathrm{mV}$ )


Figure 9. Total Output Error vs. Supply Voltage (VSENSE $=100 \mathrm{mV})$


Figure 10. Total Output Error vs. Supply Voltage ( $V_{\text {SENSE }}=6.25 \mathrm{mV}$ )


Figure 11. Total Output Error vs. Common-Mode Voltage


Figure 12. Total Output Error vs. Temperature


Figure 13. Gain Accuracy vs. Temperature


Figure 14. Output High Voltage $\left(V_{C C}-V_{O H}\right)$ vs. Temperature


Figure 15. PSRR vs. Frequency


Figure 16. Small Signal Gain vs. Frequency


Figure 17. ADM4073T Small Signal Transient Response


Figure 18. ADM4073F Small Signal Transient Response


Figure 19. ADM4073H Small Signal Transient Response


Figure 20. ADM4073T Large Signal Transient Response


Figure 21. ADM4073F Large Signal Transient Response


Figure 22. ADM4073H Large Signal Transient Response


Figure 23. ADM4073T Overdrive Response


Figure 24. ADM4073T Start-Up Delay

## ADM4073

## THEORY OF OPERATION

The current from the source flows through Rsense, which generates a voltage drop, VSENsE, across the RS+ and RS- terminals of the sense amplifier. The Input Stage Amplifier A1 regulates its inputs to be equal, thereby shunting a current proportional to $\mathrm{V}_{\text {SENSE }} / \mathrm{R}_{\mathrm{G1}}$ to the output current mirror. This current is then multiplied by a gain factor of $b$ in the output stage current mirror and flows through $\mathrm{R}_{\mathrm{GD}}$ to generate $\mathrm{V}_{\text {out }}$. Therefore, $\mathrm{V}_{\text {out }}$ is related to $\mathrm{V}_{\text {SENSE }}$ by the ratio of $\mathrm{R}_{\mathrm{G} 1}$ to $\mathrm{R}_{\mathrm{GD}}$ and the current gain of $b$.

$$
V_{\text {OUT }}=A_{V} \times V_{\text {SENSE }}
$$

where:
$A_{V}=R_{G D} / R_{G 1} \times b$
$A_{v}$ is equal to different voltages depending upon the model of the device.

- $20 \mathrm{~V} / \mathrm{V}$ for ADM4073T.
- $50 \mathrm{~V} / \mathrm{V}$ for ADM4073F.
- $100 \mathrm{~V} / \mathrm{V}$ for ADM4073H.


Figure 25. Functional Block Diagram

## $R_{\text {sense }}$

The ADM4073 has the ability to sense a wide variety of currents by selecting a particular sense resistor. Select a suitable output voltage for full-scale current, such as 10 V for 10 A . Then, select a gain model that gives the most efficient use of the sense voltage range ( 150 mV max).
In the example above, using the ADM4073H (gain of 100) gives an output voltage of 10 V when the sense voltage is 100 mV . Use the following equation to determine what value of sense resistor gives 100 mV with 10 A flowing through it:

$$
\begin{aligned}
& R_{\text {SENSE }}=100 \mathrm{mV} / 10 \mathrm{~A} \\
& R_{\text {SENSE }}=10 \mathrm{~m} \Omega \\
& V_{\text {OUT }}=\left(I_{\text {LOAD }} \times R_{\text {SENSE }}\right) \times A_{V}
\end{aligned}
$$

To measure lower currents accurately, use as large a sense resistor as possible to utilize the higher end of the sense voltage range. This reduces the effects of the offset voltage errors in the internal amplifier.
When currents are very large, it is important to take the $I^{2} R$ power losses across the sense resistor into account. If the sense resistor's rated power dissipation is not sufficient, its value can drift, giving an inaccurate output voltage or it could fail altogether. This, in turn, causes the voltage across the RS+ and RSpins to exceed the absolute maximum ratings.

If the monitored supply rail has a large amplitude high frequency component, choose a sense resistor with low inductance.


Figure 26. Using PCB Trace for Current Sensing

## OUTPUT (OUT)

The output stage of the ADM4073 is a current source driving a pull-down resistance. To ensure optimum accuracy, care must be taken not to load this output externally. To minimize output errors, ensure OUT is connected to a high impedance input stage. If this is not possible, output buffering is recommended.

The percent error introduced by output loading is determined with the following formula:

$$
\% \text { Error }=100\left(1-R_{\text {LOAD }} /\left(R_{\text {OUT_INT }^{\prime} I N T}+R_{\text {LOAD }}\right)\right)
$$

where:
$R_{\text {LOAD }}$ is the external load applied to OUT.
$R_{\text {out_INT }}$ is the internal output resistance ( $12 \mathrm{k} \Omega$ ).

## OUTLINE DIMENSIONS



Figure 27. 6-Lead Small Outline Transistor Package [SOT-23] (RJ-6)
Dimensions shown in millimeters

| Model | Gain | Temperature Range | Package Description | Package Option | Branding |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ADM4073TWRJZ-REEL71 | 20 | $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ | 6-Lead SOT-23 | RJ-6 | M2E |
| ADM4073FWRJZ-REEL7 ${ }^{1}$ | 50 | $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ | 6-Lead SOT-23 | RJ-6 | M2C |
| ADM4073HWRJZ-REEL7 ${ }^{1}$ | 100 | $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ | 6-Lead SOT-23 | RJ-6 | M2D |
| ADM4073WFWRJZ-RL7 ${ }^{1,2}$ | 50 | $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ | 6-Lead SOT-23 | RJ-6 | M2C |

${ }^{1} Z=$ RoHS Compliant Part.
${ }^{2}$ Automotive Grade.

## ADM4073

## NOTES

## X-ON Electronics

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AD8664ARUZ-REEL EVAL-ADT7411EBZ AD650JN OP400GP ADG707BRU ADSP-21469BBC-3 AD843JN ADM694SQ AD8001AR HMC444LP4ETR HMC505LP4ETR 5962-8686101XA 5962-8851301PA 5962-89710013X 5962-9169003MXA 5962-9176404M3A 59629316401MXA 5962-9452101M2A EV1HMC1160LP5 EV1HMC305SLP4 EV1HMC306AMS10 EV1HMC544A EV1HMC557ALC4 EV1HMC6146BLC5A EV1HMC6832ALP5L EV1HMC7912LP5 EV1HMC7992LP3D EV1HMC951BLP4 EV-AD5443/46/53SDZ EV-ADF70301-433AZ EV-ADF70301-868BZ EV-ADUCM322IQSPZ EV-ADUCM322QSPZ EVAL01-HMC1048LC3B EVAL01HMC1055LP2C EVAL01-HMC1063LP3 EVAL01-HMC197B EVAL01-HMC760LC4B EVAL01-HMC829LP6GE EVAL01HMC833LP6GE EVAL01-HMC835LP6G EVAL01-HMC985LP4KE EVAL01-HMC987LP5E EVAL01-HMC988LP3E EVAL01HMC995LP5GE EVAL02-HMC1034LP6G EVAL-3CH4CHSOICEBZ EVAL-AD1871EBZ EVAL-AD5063EBZ EVAL-AD5171DBZ


[^0]:    ${ }^{1} 100 \%$ production tested at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$. Specifications over temperature limit are guaranteed by design.
    ${ }^{2}$ The sum of the gain and offset errors is the total OUT voltage error.
    ${ }^{3} 6.25 \mathrm{mV}=1 / 16^{\text {th }}$ of 100 mV full-scale sense voltage.
    ${ }^{4}$ Output settles to within $1 \%$ of final value.
    ${ }^{5}$ When overdriven, this device does not experience phase reversal.

