## FEATURES

Enhanced system-level ESD performance per IEC 61000-4-x Low power operation

5 V operation
2.0 mA per channel maximum at 0 Mbps to 2 Mbps
4.1 mA per channel maximum at 10 Mbps

36 mA per channel maximum at 90 Mbps
3.3 V operation
1.0 mA per channel maximum at 0 Mbps to 2 Mbps
2.8 mA per channel maximum at 10 Mbps

17 mA per channel maximum at 90 Mbps
Bidirectional communication
3.3 V/5 V level translation

High temperature operation: $105^{\circ} \mathrm{C}$
High data rate: dc to 90 Mbps (NRZ)
Precise timing characteristics
2 ns maximum pulse width distortion
2 ns maximum channel-to-channel matching
High common-mode transient immunity: $\mathbf{> 2 5} \mathbf{~ k V} / \mu \mathrm{s}$
Output enable function
16-lead SOIC wide body, RoHS-compliant package
Safety and regulatory approvals
UL recognition: $\mathbf{2 5 0 0}$ V rms for 1 minute per UL 1577
CSA Component Acceptance Notice 5A
VDE Certificate of Conformity
DIN V VDE V 0884-10 (VDE V 0884-10): 2006-12
VIORM = 560 V peak
CQC Certification per GB4943.1-2011

## APPLICATIONS

General-purpose multichannel isolation
SPI interface/data converter isolation
RS-232/RS-422/RS-485 transceivers
Industrial field bus isolation

## GENERAL DESCRIPTION

The ADuM3300/ADuM3301 ${ }^{1}$ are 3-channel digital isolators based on the Analog Devices, Inc., iCoupler technology. Combining high speed CMOS and monolithic air core transformer technology, these isolation components provide outstanding performance characteristics superior to alternatives, such as optocoupler devices.
$i$ Coupler devices remove the design difficulties commonly associated with optocouplers. Typical optocoupler concerns regarding uncertain current transfer ratios, nonlinear transfer functions, and temperature and lifetime effects are eliminated with the simple $i$ Coupler digital interfaces and stable performance characteristics. The need for external drivers and other discrete components is eliminated with these $i$ Coupler products.
Furthermore, iCoupler devices consume one-tenth to one-sixth the power of optocouplers at comparable signal data rates.

The ADuM3300/ADuM3301 isolators provide three independent isolation channels in a variety of channel configurations and data rates (see the Ordering Guide). All models operate with the supply voltage on either side ranging from 3.3 V to 5.5 V , providing compatibility with lower voltage systems as well as enabling a voltage translation functionality across the isolation barrier. The ADuM3300/ADuM3301 isolators have a patented refresh feature that ensures dc correctness in the absence of input logic transitions and during power-up/power-down conditions.
In comparison to ADuM1300/ADuM1301 isolators, ADuM3300/ ADuM3301 isolators contain various circuit and layout changes to provide increased capability relative to system-level IEC 61000-4-x testing (ESD, burst, and surge). The precise capability in these tests for either the ADuM1300/ADuM1301 or ADuM3300/ADuM3301 products is strongly determined by the design and layout of the user's system.

## FUNCTIONAL BLOCK DIAGRAMS



Figure 1. ADuM3300 Functional Block Diagram

[^0]
## Rev. E

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Figure 2. ADuM3301 Functional Block Diagram

[^1]
## ADuM3300/ADuM3301

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## SPECIFICATIONS

## ELECTRICAL CHARACTERISTICS—5 V OPERATION

All voltages are relative to their respective ground. $4.5 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD} 1} \leq 5.5 \mathrm{~V}, 4.5 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD} 2} \leq 5.5 \mathrm{~V}$; all minimum/maximum specifications apply over the entire recommended operation range, unless otherwise noted; all typical specifications are at $T_{A}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD} 1}=\mathrm{V}_{\mathrm{DD} 2}=5 \mathrm{~V}$.

Table 1.

| Parameter | Symbol | Min | Typ | Max | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DC SPECIFICATIONS |  |  |  |  |  |  |
| Input Supply Current per Channel, Quiescent | IDDI (0) |  | 0.66 | 0.97 | mA |  |
| Output Supply Current per Channel, Quiescent | IDDo (Q) |  | 0.39 | 0.55 | mA |  |
| ADuM3300, Total Supply Current, Four Channels ${ }^{1}$ DC to 2 Mbps |  |  |  |  |  |  |
| $\mathrm{V}_{\mathrm{DD} 1}$ Supply Current | IDD1 (Q) |  | 2.4 | 3.3 | mA | DC to 1 MHz logic signal freq. |
| $V_{\text {DD2 } 2}$ Supply Current | ldD2 (Q) |  | 1.1 | 2.1 | mA | DC to 1 MHz logic signal freq. |
| 10 Mbps (BRW and CRW Grades Only) |  |  |  |  |  |  |
| $\mathrm{V}_{\text {DD } 1}$ Supply Current | $\operatorname{ldD1}(10)$ |  | 7.0 | 8.1 | mA | 5 MHz logic signal freq. |
| $V_{\text {DD2 }}$ Supply Current | IDD2 (10) |  | 2.7 | 3.6 | mA | 5 MHz logic signal freq. |
| 90 Mbps (CRW Grade Only) |  |  |  |  |  |  |
| $V_{\text {DD } 1}$ Supply Current | IDD1 (90) |  | 54 | 77 | mA | 45 MHz logic signal freq. |
| $V_{\text {DD2 }}$ Supply Current | IDD2 (90) |  | 15 | 31 | mA | 45 MHz logic signal freq. |
| ADuM3301, Total Supply Current, Four Channels ${ }^{1}$ |  |  |  |  |  |  |
| $\mathrm{V}_{\text {DD } 1}$ Supply Current | lodi (Q) |  | 2.0 | 3.1 | mA | DC to 1 MHz logic signal freq. |
| $V_{\text {DD2 }}$ Supply Current | IDD2 (Q) |  | 1.6 | 2.3 | mA | DC to 1 MHz logic signal freq. |
| 10 Mbps (BRW and CRW Grades Only) |  |  |  |  |  |  |
| $V_{\text {DD } 1}$ Supply Current | IDD1 (10) |  | 5.5 | 6.9 | mA | 5 MHz logic signal freq. |
| $V_{\text {DD2 }}$ Supply Current | ldD2 (10) |  | 3.9 | 5.4 | mA | 5 MHz logic signal freq. |
| 90 Mbps (CRW Grade Only) |  |  |  |  |  |  |
| $V_{\text {DD1 }}$ Supply Current | IDD1 (90) |  | 41 | 57 | mA | 45 MHz logic signal freq. |
| $V_{\text {DD2 }}$ Supply Current | IDD2 (90) |  | 28 | 41 | mA | 45 MHz logic signal freq. |
| For All Models |  |  |  |  |  |  |
| Input Currents | lia, lis, lic, $\mathrm{I}_{\mathrm{D}}, \mathrm{I}_{\mathrm{E} 1}, \mathrm{I}_{\mathrm{E} 2}$ | -10 | +0.01 | +10 | $\mu \mathrm{A}$ | $\begin{aligned} & 0 V \leq V_{I A}, V_{I B}, V_{1 C}, V_{I D} \leq V_{D D 1} \text { or } V_{D D 2}, \\ & 0 V \leq V_{E 1}, V_{E 2} \leq V_{D D 1} \text { or } V_{D D 2} \end{aligned}$ |
| Logic High Input Threshold | $\mathrm{V}_{\text {IH, }} \mathrm{V}_{\text {EH }}$ | 2.0 |  |  | V |  |
| Logic Low Input Threshold | $\mathrm{V}_{\text {LI, }} \mathrm{V}_{\text {EL }}$ |  |  | 0.8 | V |  |
| Logic High Output Voltages | Vоан, Vовн, Vосн, Vодн | (VDD1 or $\left.V_{D D 2}\right)-0.1$ | 5.0 |  | V | $\mathrm{l}_{\mathrm{ox}}=-20 \mu \mathrm{~A}, \mathrm{~V}_{1 \mathrm{x}}=\mathrm{V}_{1 \times \mathrm{H}}$ |
|  |  | ( $V_{D D 1}$ or $\left.V_{D D 2}\right)-0.4$ | 4.8 |  | V | $\mathrm{l}_{\mathrm{ox}}=-3.2 \mathrm{~mA}, \mathrm{~V}_{\mathrm{lx}}=\mathrm{V}_{\text {IxH }}$ |
| Logic Low Output Voltages | $\mathrm{V}_{\text {oal }} \mathrm{V}_{\text {obl, }}$ |  | 0.0 | 0.1 | V | $\mathrm{l}_{\mathrm{lx}}=20 \mu \mathrm{~A}, \mathrm{~V}_{\text {Ix }}=\mathrm{V}_{\text {lxL }}$ |
|  | Vocl, VodL |  | 0.04 | 0.1 | V | $\mathrm{l}_{\mathrm{lx}}=400 \mu \mathrm{~A}, \mathrm{~V}_{1 \mathrm{l}}=\mathrm{V}_{1 \mathrm{~lx}}$ |
|  |  |  | 0.2 | 0.4 | V | $\mathrm{l}_{\mathrm{ox}}=3.2 \mathrm{~mA}, \mathrm{~V}_{\text {lx }}=\mathrm{V}_{\text {IxL }}$ |
| SWITCHING SPECIFICATIONS |  |  |  |  |  |  |
| ADuM3300ARWZ/ADuM3301ARWZ |  |  |  |  |  |  |
| Minimum Pulse Width ${ }^{2}$ | PW |  |  | 1000 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Maximum Data Rate ${ }^{3}$ |  | 1 |  |  | Mbps | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Propagation Delay ${ }^{4}$ | $\mathrm{t}_{\text {PHL, }} \mathrm{t}_{\text {PLH }}$ | 50 | 65 | 100 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Pulse Width Distortion, \|tpLH $-\left.\mathrm{t}_{\text {PHL }}\right\|^{4}$ | PWD |  |  | 40 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Propagation Delay Skew ${ }^{5}$ | tpsk |  |  | 50 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Channel-to-Channel Matching ${ }^{6}$ | tPskco/od |  |  | 50 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |


| Parameter | Symbol | Min | Typ | Max | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADuM3300BRWZ/ADuM3301BRWZ |  |  |  |  |  |  |
| Minimum Pulse Width ${ }^{2}$ | PW |  |  | 100 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Maximum Data Rate ${ }^{3}$ |  | 10 |  |  | Mbps | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Propagation Delay ${ }^{4}$ | $\mathrm{t}_{\text {PHL, }} \mathrm{t}_{\text {PLH }}$ | 20 | 32 | 50 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Pulse Width Distortion, $\left\|\mathrm{t}_{\text {PLH }}-\mathrm{t}_{\text {PHL }}\right\|^{4}$ | PWD |  |  | 3 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Change vs. Temperature |  |  | 5 |  | $\mathrm{ps} /{ }^{\circ} \mathrm{C}$ | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Propagation Delay Skew ${ }^{5}$ | tPsk |  |  | 15 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Channel-to-Channel Matching, Codirectional Channels ${ }^{6}$ | tPSKCD |  |  | 3 | ns | $C_{L}=15 \mathrm{pF}$, CMOS signal levels |
| Channel-to-Channel Matching, Opposing-Directional Channels ${ }^{6}$ | tPskod |  |  | 6 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| ADuM3300CRWZ/ADuM3301CRWZ |  |  |  |  |  |  |
| Minimum Pulse Width ${ }^{2}$ | PW |  | 8.3 | 11.1 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Maximum Data Rate ${ }^{3}$ |  | 90 | 120 |  | Mbps | $C_{L}=15 \mathrm{pF}$, CMOS signal levels |
| Propagation Delay ${ }^{4}$ | $\mathrm{tPHL} \mathrm{t}_{\text {PLL }}$ | 18 | 27 | 32 | ns | $C_{L}=15 \mathrm{pF}$, CMOS signal levels |
| Pulse Width Distortion, $\left\|\mathrm{t}_{\text {PLH }}-\mathrm{t}_{\text {PHL }}\right\|^{4}$ | PWD |  | 0.5 | 2 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Change vs. Temperature |  |  | 3 |  | $\mathrm{ps} /{ }^{\circ} \mathrm{C}$ | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Propagation Delay Skew ${ }^{5}$ | tPSK |  |  | 10 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Channel-to-Channel Matching, Codirectional Channels ${ }^{6}$ | tPSKCD |  |  | 2 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Channel-to-Channel Matching, Opposing-Directional Channels ${ }^{6}$ | tPSKOD |  |  | 5 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| For All Models |  |  |  |  |  |  |
| Output Disable Propagation Delay (High/Low-to-High Impedance) | $\mathrm{t}_{\text {PHZ }}, \mathrm{t}_{\text {PLH }}$ |  | 6 | 8 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Output Enable Propagation Delay (High Impedance-to-High/Low) | $\mathrm{t}_{\text {PZH, }} \mathrm{t}_{\text {PZL }}$ |  | 6 | 8 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Output Rise/Fall Time (10\% to 90\%) | $\mathrm{t}_{\mathrm{R}} / \mathrm{t}_{\mathrm{F}}$ |  | 2.5 |  | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Common-Mode Transient Immunity at Logic High Output ${ }^{7}$ | \|CM ${ }^{\text {\| }}$ | 25 | 35 |  | $\mathrm{kV} / \mu \mathrm{s}$ | $\begin{aligned} & \mathrm{V}_{\mathrm{Ix}}=\mathrm{V}_{\mathrm{DD} 1} \text { or } \mathrm{V}_{\mathrm{DD} 2}, \mathrm{~V}_{\mathrm{CM}}=1000 \mathrm{~V}, \\ & \text { transient magnitude }=800 \mathrm{~V} \end{aligned}$ |
| Common-Mode Transient Immunity at Logic Low Output ${ }^{7}$ | $\mid C M$ \| | 25 | 35 |  | $\mathrm{kV} / \mu \mathrm{s}$ | $\mathrm{V}_{\mathrm{lx}}=0 \mathrm{~V}, \mathrm{~V}_{\mathrm{CM}}=1000 \mathrm{~V} \text {, }$ <br> transient magnitude $=800 \mathrm{~V}$ |
| Refresh Rate | $\mathrm{fr}_{\mathrm{r}}$ |  | 1.2 |  | Mbps |  |
| Input Dynamic Supply Current per Channel ${ }^{8}$ | l DII (D) |  | 0.20 |  | mA/Mbps |  |
| Output Dynamic Supply Current per Channel ${ }^{8}$ | IDDO (D) |  | 0.05 |  | mA/Mbps |  |

${ }^{1}$ The supply current values for all four channels are combined when running at identical data rates. Output supply current values are specified with no output load present. The supply current associated with an individual channel operating at a given data rate can be calculated as described in the Power Consumption section. See Figure 6 through Figure 8 for information on per-channel supply current as a function of data rate for unloaded and loaded conditions. See Figure 9 through Figure 12 for total $V_{D D 1}$ and $V_{D D 2}$ supply currents as a function of data rate for ADuM3300/ADuM3301 channel configurations.
${ }^{2}$ The minimum pulse width is the shortest pulse width at which the specified pulse width distortion is guaranteed.
${ }^{3}$ The maximum data rate is the fastest data rate at which the specified pulse width distortion is guaranteed.
${ }^{4} t_{\text {PHL }}$ propagation delay is measured from the $50 \%$ level of the falling edge of the $V_{1 X}$ signal to the $50 \%$ level of the falling edge of the $V_{O x}$ signal. $t_{\text {PLH }}$ propagation delay is measured from the $50 \%$ level of the rising edge of the $V_{1 x}$ signal to the $50 \%$ level of the rising edge of the $V_{\text {ox }}$ signal.
${ }^{5} t_{\text {PSK }}$ is the magnitude of the worst-case difference in $t_{\text {PHL }}$ or $t_{\text {PLH }}$ that is measured between units at the same operating temperature, supply voltages, and output load within the recommended operating conditions.
${ }^{6}$ Codirectional channel-to-channel matching is the absolute value of the difference in propagation delays between any two channels with inputs on the same side of the isolation barrier. Opposing-directional channel-to-channel matching is the absolute value of the difference in propagation delays between any two channels with inputs on opposing sides of the isolation barrier.
${ }^{7} \mathrm{CM}_{\mathrm{H}}$ is the maximum common-mode voltage slew rate that can be sustained while maintaining $\mathrm{V}_{\mathrm{O}}>0.8 \mathrm{~V}_{\mathrm{DD} 2}$. CML is the maximum common-mode voltage slew rate that can be sustained while maintaining $\mathrm{V}_{\mathrm{o}}<0.8 \mathrm{~V}$. The common-mode voltage slew rates apply to both rising and falling common-mode voltage edges. The transient magnitude is the range over which the common mode is slewed.
${ }^{8}$ Dynamic supply current is the incremental amount of supply current required for a 1 Mbps increase in signal data rate. See Figure 6 through Figure 8 for information on per-channel supply current for unloaded and loaded conditions. See the Power Consumption section for guidance on calculating the per-channel supply current for a given data rate.

## ELECTRICAL CHARACTERISTICS—3.3 V OPERATION

All voltages are relative to their respective ground. $3.0 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD} 1} \leq 3.6 \mathrm{~V}, 3.0 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD} 2} \leq 3.6 \mathrm{~V}$; all minimum/maximum specifications apply over the entire recommended operation range, unless otherwise noted; all typical specifications are at $T_{A}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD} 1}=\mathrm{V}_{\mathrm{DD} 2}=3.3 \mathrm{~V}$.

Table 2.

| Parameter | Symbol | Min | Typ | Max | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DC SPECIFICATIONS |  |  |  |  |  |  |
| Input Supply Current per Channel, Quiescent | IDDI (0) |  | 0.37 | 0.57 | mA |  |
| Output Supply Current per Channel, Quiescent | IDDo (Q) |  | 0.25 | 0.37 | mA |  |
| ADuM3300, Total Supply Current, Four Channels ${ }^{1}$ DC to 2 Mbps |  |  |  |  |  |  |
|  |  |  |  |  |  |  |
| $V_{\text {DD1 }}$ Supply Current | $\mathrm{IDD1}$ (Q) |  | 1.4 | 1.9 | mA | DC to 1 MHz logic signal freq. |
| $\mathrm{V}_{\mathrm{DD} 2}$ Supply Current | $\mathrm{ldD2}$ (Q) |  | 0.7 | 1.2 | mA | DC to 1 MHz logic signal freq. |
| 10 Mbps (BRW and CRW Grades Only) |  |  |  |  |  |  |
| $V_{\text {DDI }}$ Supply Current | IDD1 (10) |  | 3.8 | 5.3 | mA | 5 MHz logic signal freq. |
| $V_{\text {DD2 }}$ Supply Current | IDD2 (10) |  | 1.5 | 2.1 | mA | 5 MHz logic signal freq. |
| 90 Mbps (CRW Grade Only) |  |  |  |  |  |  |
| $V_{\text {DD } 1}$ Supply Current | IDD1 (90) |  | 28 | 41 | mA | 45 MHz logic signal freq. |
| $V_{\text {DD2 }}$ Supply Current | IDD2 (90) |  | 8.2 | 11 | mA | 45 MHz logic signal freq. |
| ADuM3301, Total Supply Current, Four Channels ${ }^{1}$ |  |  |  |  |  |  |
| DC to 2 Mbps |  |  |  |  |  |  |
| $V_{\text {DD } 1}$ Supply Current | $\mathrm{ldD1}$ (Q) |  | 1.1 | 1.6 | mA | DC to 1 MHz logic signal freq. |
| $V_{\text {DD2 }}$ Supply Current | $\mathrm{ldD2}$ (Q) |  | 0.9 | 1.4 | mA | DC to 1 MHz logic signal freq. |
| 10 Mbps (BRW and CRW Grades Only) |  |  |  |  |  |  |
| $V_{\text {DD } 1}$ Supply Current | IDD1 (10) |  | 3.0 | 4.1 | mA | 5 MHz logic signal freq. |
| $\mathrm{V}_{\mathrm{DD} 2}$ Supply Current | $\mathrm{ldD2}$ (10) |  | 2.2 | 2.9 | mA | 5 MHz logic signal freq. |
| 90 Mbps (CRW Grade Only) |  |  |  |  |  |  |
| $V_{\text {DD } 1}$ Supply Current | IDD1 (90) |  | 22 | 31 | mA | 45 MHz logic signal freq. |
| $\mathrm{V}_{\text {DD2 }}$ Supply Current | ldD2 (90) |  | 15 | 21 | mA | 45 MHz logic signal freq. |
| For All Models |  |  |  |  |  |  |
| Input Currents | $\mathrm{I}_{\mathrm{I}}$, I lis, I Ic, $\mathrm{I}_{\mathrm{ID},} \mathrm{I}_{\mathrm{E} 1}, \mathrm{I}_{\mathrm{E} 2}$ | -10 | +0.01 | +10 | $\mu \mathrm{A}$ | $\begin{aligned} & 0 \mathrm{~V} \leq \mathrm{V}_{\mathrm{IA},}, \mathrm{~V}_{\mathrm{IB},} \mathrm{~V}_{\mathrm{VC},}, \mathrm{~V}_{1 \mathrm{D}} \leq \mathrm{V}_{\mathrm{DD} 1} \text { or } \mathrm{V}_{\mathrm{DD} 2,}, \\ & 0 \mathrm{~V} \leq \mathrm{V}_{\mathrm{E} 1}, \mathrm{~V}_{\mathrm{E} 2} \leq \mathrm{V}_{\mathrm{DD} 1} \text { or } \mathrm{V}_{\mathrm{DD} 2} \end{aligned}$ |
| Logic High Input Threshold | $\mathrm{V}_{\text {IH }}, \mathrm{V}_{\text {EH }}$ | 1.6 |  |  | V |  |
| Logic Low Input Threshold | $\mathrm{V}_{\text {LL }}, \mathrm{V}_{\text {EL }}$ |  |  | 0.4 | V |  |
| Logic High Output Voltages | Voaн, $\mathrm{V}_{\text {овн, }}$ Vосн, $V_{\text {одн }}$ | (VDD $\left.V_{D D 2}\right)-0.1$ | 3.0 |  | V | $\mathrm{l}_{\mathrm{ox}}=-20 \mu \mathrm{~A}, \mathrm{~V}_{1 \mathrm{x}}=\mathrm{V}_{1 \times \mathrm{H}}$ |
|  |  | $\begin{aligned} & \left(V_{D D 1}\right. \text { or } \\ & \left.V_{D D 2}\right)-0.4 \end{aligned}$ | 2.8 |  | V | $\mathrm{l}_{\mathrm{ox}}=-3.2 \mathrm{~mA}, \mathrm{~V}_{1 \mathrm{x}}=\mathrm{V}_{1 \mathrm{xH}}$ |
| Logic Low Output Voltages | Voal, $\mathrm{V}_{\text {obl, }}$ |  | 0.0 | 0.1 | V | $\mathrm{l}_{\mathrm{lox}}=20 \mu \mathrm{~A}, \mathrm{~V}_{1 \mathrm{l}}=\mathrm{V}_{\text {IxL }}$ |
|  | Vocl, $\mathrm{V}_{\text {OdL }}$ |  | 0.04 | 0.1 | V | $\mathrm{l}_{\text {lox }}=400 \mu \mathrm{~A}, \mathrm{~V}_{\text {lx }}=\mathrm{V}_{\text {lxL }}$ |
|  |  |  | 0.2 | 0.4 | V | $\mathrm{lox}^{\text {a }}=3.2 \mathrm{~mA}, \mathrm{~V}_{\text {lx }}=\mathrm{V}_{\text {IxL }}$ |
| SWITCHING SPECIFICATIONS |  |  |  |  |  |  |
| ADuM3300ARWZ/ADuM3301ARWZ |  |  |  |  |  |  |
| Minimum Pulse Width ${ }^{2}$ | PW |  |  | 1000 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Maximum Data Rate ${ }^{3}$ |  | 1 |  |  | Mbps | $C_{L}=15 \mathrm{pF}$, CMOS signal levels |
| Propagation Delay ${ }^{4}$ | $\mathrm{t}_{\text {PHL, }} \mathrm{tPLH}$ | 50 | 75 | 100 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Pulse Width Distortion, \|tpLH - ${ }_{\text {PHLL }}{ }^{4}$ | PWD |  |  | 40 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Propagation Delay Skew ${ }^{5}$ | tpsk |  |  | 50 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Channel-to-Channel Matching ${ }^{6}$ | tPskco/od |  |  | 50 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |


| Parameter | Symbol | Min | Typ | Max | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADuM3300BRWZ/ADuM3301BRWZ |  |  |  |  |  |  |
| Minimum Pulse Width ${ }^{2}$ | PW |  |  | 100 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Maximum Data Rate ${ }^{3}$ |  | 10 |  |  | Mbps | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Propagation Delay ${ }^{4}$ | $\mathrm{t}_{\text {PHL, }} \mathrm{t}_{\text {PLL }}$ | 20 | 38 | 50 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Pulse Width Distortion, $\left\|t_{\text {PLL }}-t_{\text {PHL }}\right\|^{4}$ | PWD |  |  | 3 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Change vs. Temperature |  |  | 5 |  | ps/ $/{ }^{\circ} \mathrm{C}$ | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Propagation Delay Skew ${ }^{5}$ | tpsk |  |  | 22 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Channel-to-Channel Matching, Codirectional Channels ${ }^{6}$ | tPSKCD |  |  | 3 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Channel-to-Channel Matching, Opposing-Directional Channels ${ }^{6}$ | tPSKOD |  |  | 6 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| ADuM3300CRWZ/ADuM3301CRWZ |  |  |  |  |  |  |
| Minimum Pulse Width ${ }^{2}$ | PW |  | 8.3 | 11.1 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Maximum Data Rate ${ }^{3}$ |  | 90 | 120 |  | Mbps | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Propagation Delay ${ }^{4}$ | tPHL, tPLH | 20 | 34 | 45 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Pulse Width Distortion, $\left\|\mathrm{t}_{\text {PLH }}-\mathrm{t}_{\text {PHL }}\right\|^{4}$ | PWD |  | 0.5 | 2 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Change vs. Temperature |  |  | 3 |  | $\mathrm{ps} /{ }^{\circ} \mathrm{C}$ | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Propagation Delay Skew ${ }^{5}$ | tpsk |  |  | 16 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Channel-to-Channel Matching, Codirectional Channels ${ }^{6}$ | $\mathrm{t}_{\text {PSKCD }}$ |  |  | 2 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Channel-to-Channel Matching, Opposing-Directional Channels ${ }^{6}$ | tPSKOD |  |  | 5 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| For All Models |  |  |  |  |  |  |
| Output Disable Propagation Delay (High/Low-to-High Impedance) | $\mathrm{t}_{\text {PHZ, }} \mathrm{t}_{\text {PLH }}$ |  | 6 | 8 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Output Enable Propagation Delay (High Impedance-to-High/Low) | $\mathrm{t}_{\text {PZH, }} \mathrm{t}_{\text {PZL }}$ |  | 6 | 8 | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Output Rise/Fall Time (10\% to 90\%) | $\mathrm{t}_{\mathrm{R}} / \mathrm{t}_{\mathrm{F}}$ |  | 3 |  | ns | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Common-Mode Transient Immunity at Logic High Output ${ }^{7}$ | \|CM ${ }^{\text {\| }}$ | 25 | 35 |  | $\mathrm{kV} / \mu \mathrm{s}$ | $\begin{aligned} & \mathrm{V}_{\mathrm{Ix}}=\mathrm{V}_{\mathrm{DD} 1} \text { or } \mathrm{V}_{\mathrm{DD} 2}, \mathrm{~V}_{\mathrm{CM}}=1000 \mathrm{~V}, \\ & \text { transient magnitude }=800 \mathrm{~V} \end{aligned}$ |
| Common-Mode Transient Immunity at Logic Low Output ${ }^{7}$ | \|CML| | 25 | 35 |  | $\mathrm{kV} / \mu \mathrm{s}$ | $\begin{aligned} & \mathrm{V}_{1 \mathrm{x}}=0 \mathrm{~V}, \mathrm{~V}_{\mathrm{CM}}=1000 \mathrm{~V}, \\ & \text { transient magnitude }=800 \mathrm{~V} \end{aligned}$ |
| Refresh Rate | $\mathrm{fr}_{\mathrm{r}}$ |  | 1.1 |  | Mbps |  |
| Input Dynamic Supply Current per Channel ${ }^{8}$ | IDDI (D) |  | 0.10 |  | mA/Mbps |  |
| Output Dynamic Supply Current per Channel ${ }^{8}$ | IDDO (D) |  | 0.03 |  | mA/Mbps |  |

${ }^{1}$ The supply current values for all four channels are combined when running at identical data rates. Output supply current values are specified with no output load present. The supply current associated with an individual channel operating at a given data rate can be calculated as described in the Power Consumption section. See Figure 6 through Figure 8 for information on per-channel supply current as a function of data rate for unloaded and loaded conditions. See Figure 9 through Figure 12 for total $\mathrm{V}_{\mathrm{DD} 1}$ and $\mathrm{V}_{\mathrm{DD} 2}$ supply currents as a function of data rate for ADuM3300/ADuM3301 channel configurations.
${ }^{2}$ The minimum pulse width is the shortest pulse width at which the specified pulse width distortion is guaranteed.
${ }^{3}$ The maximum data rate is the fastest data rate at which the specified pulse width distortion is guaranteed.
${ }^{4} t_{\text {PHL }}$ propagation delay is measured from the $50 \%$ level of the falling edge of the $V_{\mathrm{IX}}$ signal to the $50 \%$ level of the falling edge of the $V_{O x}$ signal. $t_{\text {PLH }}$ propagation delay is measured from the $50 \%$ level of the rising edge of the $V_{1 x}$ signal to the $50 \%$ level of the rising edge of the $V_{\text {ox }}$ signal.
${ }^{5} t_{\text {PSK }}$ is the magnitude of the worst-case difference in $t_{P H L}$ or $t_{P L H}$ that is measured between units at the same operating temperature, supply voltages, and output load within the recommended operating conditions.
${ }^{6}$ Codirectional channel-to-channel matching is the absolute value of the difference in propagation delays between any two channels with inputs on the same side of the isolation barrier. Opposing-directional channel-to-channel matching is the absolute value of the difference in propagation delays between any two channels with inputs on opposing sides of the isolation barrier.
${ }^{7} \mathrm{CM}_{H}$ is the maximum common-mode voltage slew rate that can be sustained while maintaining $\mathrm{V}_{\mathrm{O}}>0.8 \mathrm{~V}_{\mathrm{DD} 2}$. CML is the maximum common-mode voltage slew rate that can be sustained while maintaining $\mathrm{V}_{\mathrm{o}}<0.8 \mathrm{~V}$. The common-mode voltage slew rates apply to both rising and falling common-mode voltage edges. The transient magnitude is the range over which the common mode is slewed.
${ }^{8}$ Dynamic supply current is the incremental amount of supply current required for a 1 Mbps increase in signal data rate. See Figure 6 through Figure 8 for information on per-channel supply current for unloaded and loaded conditions. See the Power Consumption section for guidance on calculating the per-channel supply current for a given data rate.

## ELECTRICAL CHARACTERISTICS—MIXED 5 V/3.3 V OR 3.3 V/5 V OPERATION

All voltages are relative to their respective ground. $5 \mathrm{~V} / 3.3 \mathrm{~V}$ operation: $4.5 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD} 1} \leq 5.5 \mathrm{~V}, 3.0 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD} 2} \leq 3.6 \mathrm{~V} .3 .3 \mathrm{~V} / 5 \mathrm{~V}$ operation: $3.0 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD} 1} \leq 3.6 \mathrm{~V}, 4.5 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD} 2} \leq 5.5 \mathrm{~V}$. All minimum/maximum specifications apply over the entire recommended operation range, unless otherwise noted. All typical specifications are at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C} ; \mathrm{V}_{\mathrm{DD} 1}=3.3 \mathrm{~V}, \mathrm{~V}_{\mathrm{DD} 2}=5 \mathrm{~V}$ or $\mathrm{V}_{\mathrm{DD} 1}=5 \mathrm{~V}, \mathrm{~V}_{\mathrm{DD} 2}=3.3 \mathrm{~V}$.

Table 3.


## ADuM3300/ADuM3301

| Parameter | Symbol | Min | Typ | Max | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 90 Mbps (CRW Grade Only) |  |  |  |  |  |  |
| VDD1 Supply Current | l DD1 (90) |  |  |  |  |  |
| $5 \mathrm{~V} / 3 \mathrm{~V}$ Operation |  |  | 41 | 57 | mA | 45 MHz logic signal freq. |
| $3 \mathrm{~V} / 5 \mathrm{~V}$ Operation |  |  | 22 | 31 | mA | 45 MHz logic signal freq. |
| $\mathrm{V}_{\text {DD2 }}$ Supply Current | IDD2 (90) |  |  |  |  |  |
| $5 \mathrm{~V} / 3 \mathrm{~V}$ Operation |  |  | 15 | 21 | mA | 45 MHz logic signal freq. |
| $3 \mathrm{~V} / 5 \mathrm{~V}$ Operation |  |  | 28 | 41 | mA | 45 MHz logic signal freq. |
| For All Models |  |  |  |  |  |  |
| Input Currents | $I_{I A}, I_{I B}, I_{I C}$, $\mathrm{I}_{\mathrm{ID}}, \mathrm{I}_{\mathrm{E} 1}, \mathrm{I}_{\mathrm{E} 2}$ | -10 | +0.01 | +10 | $\mu \mathrm{A}$ | $\begin{aligned} & 0 \mathrm{~V} \leq \mathrm{V}_{\mathrm{IA}}, \mathrm{~V}_{\mathrm{IB},}, \mathrm{~V}_{\mathrm{CC}}, \mathrm{~V}_{\mathrm{ID}} \leq \mathrm{V}_{\mathrm{DD} 1} \text { or } \mathrm{V}_{\mathrm{DD} 2,}, \\ & 0 \mathrm{~V} \leq \mathrm{V}_{\mathrm{E} 1}, \mathrm{~V}_{\mathrm{E} 2} \leq \mathrm{V}_{\mathrm{DD} 1} \text { or } \mathrm{V}_{\mathrm{DD}} \end{aligned}$ |
| Logic High Input Threshold | $\mathrm{V}_{\mathrm{IH}}, \mathrm{V}_{\text {EH }}$ |  |  |  |  |  |
| $5 \mathrm{~V} / 3 \mathrm{~V}$ Operation |  | 2.0 |  |  | V |  |
| $3 \mathrm{~V} / 5 \mathrm{~V}$ Operation |  | 1.6 |  |  | V |  |
| Logic Low Input Threshold | VIL, $\mathrm{V}_{\text {EL }}$ |  |  |  |  |  |
| $5 \mathrm{~V} / 3 \mathrm{~V}$ Operation |  |  |  | 0.8 | V |  |
| $3 \mathrm{~V} / 5 \mathrm{~V}$ Operation |  |  |  | 0.4 | V |  |
| Logic High Output Voltages | $\mathrm{V}_{\text {OAh, }} \mathrm{V}_{\text {ObH, }}$ Voch, Vodh | $\left\lvert\, \begin{aligned} & \left(\mathrm{V}_{\mathrm{DD} 1} \mathrm{or}\right. \\ & \left.\mathrm{V}_{\mathrm{DD} 2}\right)-0.1 \end{aligned}\right.$ | $\left(V_{D D 1} \text { or } V_{D D 2}\right)$ |  | V | $\mathrm{I}_{\mathrm{Ox}}=-20 \mu \mathrm{~A}, \mathrm{~V}_{\mathrm{Ix}}=\mathrm{V}_{\mathrm{IxH}}$ |
|  |  | $\begin{aligned} & \left(V_{D D 1}\right. \text { or } \\ & \left.V_{D D 2}\right)-0.4 \end{aligned}$ | $\begin{aligned} & \left(V_{D D 1}\right. \text { or } \\ & \left.V_{D D 2}\right)-0.2 \end{aligned}$ |  | V | $\mathrm{I}_{\mathrm{ox}}=-3.2 \mathrm{~mA}, \mathrm{~V}_{\mathrm{Ix}}=\mathrm{V}_{\mathrm{IXH}}$ |
| Logic Low Output Voltages | Voal, Vobl, |  | 0.0 | 0.1 | V | $\mathrm{l}_{0 \mathrm{x}}=20 \mu \mathrm{~A}, \mathrm{~V}_{1 \mathrm{x}}=\mathrm{V}_{1 \times \mathrm{L}}$ |
|  | $\mathrm{V}_{\text {OCL }}$, $\mathrm{V}_{\text {ODL }}$ |  | 0.04 | 0.1 | V | $\mathrm{loxx}=400 \mu \mathrm{~A}, \mathrm{~V}_{\text {Ix }}=\mathrm{V}_{\text {IxL }}$ |
|  |  |  | 0.2 | 0.4 | V | $\mathrm{l}_{\mathrm{ox}}=3.2 \mathrm{~mA}, \mathrm{~V}_{\mathrm{Ix}}=\mathrm{V}_{\text {IxL }}$ |
| SWITCHING SPECIFICATIONS |  |  |  |  |  |  |
| ADuM3300ARWZ/ADuM3301ARWZ |  |  |  |  |  |  |
| Minimum Pulse Width ${ }^{2}$ | PW |  |  | 1000 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Maximum Data Rate ${ }^{3}$ |  | 1 |  |  | Mbps | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Propagation Delay ${ }^{4}$ | $\mathrm{t}_{\text {PHL, }} \mathrm{t}_{\text {PLL }}$ | 50 | 70 | 100 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Pulse Width Distortion, $\mid$ tPLH - tPHL $\left.\right\|^{4}$ | PWD |  |  | 40 | ns | $\mathrm{C}_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Propagation Delay Skew ${ }^{5}$ | tpsk |  |  | 50 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Channel-to-Channel Matching ${ }^{6}$ | tPSKCD/OD |  |  | 50 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| ADuM3300BRWZ/ADuM3301BRWZ |  |  |  |  |  |  |
| Minimum Pulse Width ${ }^{2}$ | PW |  |  | 100 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Maximum Data Rate ${ }^{3}$ |  | 10 |  |  | Mbps | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Propagation Delay ${ }^{4}$ | $\mathrm{t}_{\text {PHL, }} \mathrm{t}_{\text {PLH }}$ | 15 | 35 | 50 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Pulse Width Distortion, $\left\|\mathrm{t}_{\text {LLH }}-\mathrm{t}_{\text {PHL }}\right\|^{4}$ | PWD |  |  | 3 | ns | $\mathrm{C}_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Change vs. Temperature |  |  | 5 |  | $\mathrm{ps} /{ }^{\circ} \mathrm{C}$ | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Propagation Delay Skew ${ }^{5}$ | tpsk |  |  | 22 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Channel-to-Channel Matching, Codirectional Channels ${ }^{6}$ | tPSKCD |  |  | 3 | ns | $\mathrm{CL}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Channel-to-Channel Matching, Opposing-Directional Channels ${ }^{6}$ | $\mathrm{t}_{\text {PSKOD }}$ |  |  | 6 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| ADuM3300CRWZ/ADuM3301CRWZ |  |  |  |  |  |  |
| Minimum Pulse Width ${ }^{2}$ | PW |  | 8.3 | 11.1 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Maximum Data Rate ${ }^{3}$ |  | 90 | 120 |  | Mbps | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Propagation Delay ${ }^{4}$ | tPhL, ${ }_{\text {Plı }}$ | 20 | 30 | 40 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Pulse Width Distortion, $\left\|t_{\text {PLH }}-t_{\text {PHL }}\right\|^{4}$ | PWD |  | 0.5 | 2 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Change vs. Temperature |  |  | 3 |  | $\mathrm{ps} /{ }^{\circ} \mathrm{C}$ | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Propagation Delay Skew ${ }^{5}$ | tPSK |  |  | 14 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Channel-to-Channel Matching, Codirectional Channels ${ }^{6}$ | $\mathrm{t}_{\text {PSKCD }}$ |  |  | 2 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Channel-to-Channel Matching, Opposing-Directional Channels ${ }^{6}$ | $\mathrm{t}_{\text {PSKOD }}$ |  |  | 5 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |


| Parameter | Symbol | Min | Typ | Max | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| For All Models |  |  |  |  |  |  |
| Output Disable Propagation Delay (High/Low-to-High Impedance) | $\mathrm{t}_{\text {PHZ }} \mathrm{t}_{\text {PLH }}$ |  | 6 | 8 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Output Enable Propagation Delay (High Impedance-to-High/Low) | tpzH, tpzL |  | 6 | 8 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| Output Rise/Fall Time (10\% to 90\%) | $\mathrm{t}_{\mathrm{R}} / \mathrm{t}_{\mathrm{F}}$ |  |  |  |  | $C_{L}=15 \mathrm{pF}, \mathrm{CMOS}$ signal levels |
| $5 \mathrm{~V} / 3.3 \mathrm{~V}$ Operation |  |  | 3.0 |  | ns |  |
| $3.3 \mathrm{~V} / 5 \mathrm{~V}$ Operation |  |  | 2.5 |  | ns |  |
| Common-Mode Transient Immunity at Logic High Output ${ }^{7}$ | \|CMH| | 25 | 35 |  | kV/ $\mu \mathrm{s}$ | $\begin{aligned} & \mathrm{V}_{\mathrm{Ix}}=\mathrm{V}_{\mathrm{DD} 1} \text { or } \mathrm{V}_{\mathrm{DD} 2}, \mathrm{~V}_{\mathrm{CM}}=1000 \mathrm{~V}, \\ & \text { transient magnitude }=800 \mathrm{~V} \end{aligned}$ |
| Common-Mode Transient Immunity at Logic Low Output ${ }^{7}$ | $\left\|C M_{L}\right\|$ | 25 | 35 |  | kV/ $\mu \mathrm{s}$ | $\begin{aligned} & \mathrm{V}_{\mathrm{lx}}=0 \mathrm{~V}, \mathrm{~V}_{\mathrm{CM}}=1000 \mathrm{~V}, \\ & \text { transient magnitude }=800 \mathrm{~V} \end{aligned}$ |
| Refresh Rate | $\mathrm{fr}_{\mathrm{r}}$ |  |  |  |  |  |
| $5 \mathrm{~V} / 3.3 \mathrm{~V}$ Operation |  |  | 1.2 |  | Mbps |  |
| $3.3 \mathrm{~V} / 5 \mathrm{~V}$ Operation |  |  | 1.1 |  | Mbps |  |
| Input Dynamic Supply Current per Channel ${ }^{8}$ | $\mathrm{ldDI}(\mathrm{D})$ |  |  |  |  |  |
| $5 \mathrm{~V} / 3.3 \mathrm{~V}$ Operation |  |  | 0.20 |  | mA/Mbps |  |
| $3.3 \mathrm{~V} / 5 \mathrm{~V}$ Operation |  |  | 0.10 |  | mA/Mbps |  |
| Output Dynamic Supply Current per Channel ${ }^{8}$ | IDDO (D) |  |  |  |  |  |
| $5 \mathrm{~V} / 3.3 \mathrm{~V}$ Operation |  |  | 0.05 |  | mA/Mbps |  |
| $3.3 \mathrm{~V} / 5 \mathrm{~V}$ Operation |  |  | 0.03 |  | mA/Mbps |  |

${ }^{1}$ The supply current values for all four channels are combined when running at identical data rates. Output supply current values are specified with no output load present. The supply current associated with an individual channel operating at a given data rate can be calculated as described in the Power Consumption section. See Figure 6 through Figure 8 for information on per-channel supply current as a function of data rate for unloaded and loaded conditions. See Figure 9 through Figure 12 for total $\mathrm{V}_{\mathrm{DD} 1}$ and $\mathrm{V}_{\mathrm{DD} 2}$ supply currents as a function of data rate for ADuM3300/ADuM3301 channel configurations.
${ }^{2}$ The minimum pulse width is the shortest pulse width at which the specified pulse width distortion is guaranteed.
${ }^{3}$ The maximum data rate is the fastest data rate at which the specified pulse width distortion is guaranteed.
 measured from the $50 \%$ level of the rising edge of the $\mathrm{V}_{1 \mathrm{x}}$ signal to the $50 \%$ level of the rising edge of the $\mathrm{V}_{0 \times}$ signal.
${ }^{5} t_{\text {Psk }}$ is the magnitude of the worst-case difference in $t_{\text {PHL }}$ or $t_{\text {PLH }}$ that is measured between units at the same operating temperature, supply voltages, and output load within the recommended operating conditions.
${ }^{6}$ Codirectional channel-to-channel matching is the absolute value of the difference in propagation delays between any two channels with inputs on the same side of the isolation barrier. Opposing-directional channel-to-channel matching is the absolute value of the difference in propagation delays between any two channels with inputs on opposing sides of the isolation barrier.
${ }^{7} \mathrm{CM}_{H}$ is the maximum common-mode voltage slew rate that can be sustained while maintaining $\mathrm{V}_{\mathrm{O}}>0.8 \mathrm{~V}_{\mathrm{DD} 2}$. CML is the maximum common-mode voltage slew rate that can be sustained while maintaining $\mathrm{V}_{0}<0.8 \mathrm{~V}$. The common-mode voltage slew rates apply to both rising and falling common-mode voltage edges. The transient magnitude is the range over which the common mode is slewed.
${ }^{8}$ Dynamic supply current is the incremental amount of supply current required for a 1 Mbps increase in signal data rate. See Figure 6 through Figure 8 for information on per-channel supply current for unloaded and loaded conditions. See the Power Consumption section for guidance on calculating the per-channel supply current for a given data rate.

## ADuM3300/ADuM3301

## PACKAGE CHARACTERISTICS

Table 4.

| Parameter | Symbol | Min | Typ $\quad$ Max | Unit | Test Conditions |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Resistance (Input to Output) $)^{1}$ | $\mathrm{R}_{1-\mathrm{O}}$ | $10^{12}$ | $\Omega$ |  |  |
| Capacitance (Input to Output) $^{1}$ | $\mathrm{C}_{1-\mathrm{O}}$ | 2.2 | pF | $\mathrm{f}=1 \mathrm{MHz}$ |  |
| Input Capacitance $^{2}$ | $\mathrm{C}_{1}$ |  | 4.0 | pF |  |
| IC Junction-to-Case Thermal Resistance, Side 1 | $\theta_{\mathrm{jlI}}$ | 33 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ | Thermocouple located at |  |
| IC Junction-to-Case Thermal Resistance, Side 2 | $\theta_{\mathrm{Jco}}$ | 28 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ | center of package underside |  |

${ }^{1}$ The device is considered a 2-terminal device; Pin 1 through Pin 8 are shorted together, and Pin 9 through Pin 16 are shorted together.
${ }^{2}$ Input capacitance is from any input data pin to ground.
${ }^{2}$ Input capacitance is from any input data pin to ground.

## REGULATORY INFORMATION

The ADuM3300/ADuM3301 is approved by the organizations listed in Table 5. See Table 10 and the Insulation Lifetime section for details regarding recommended maximum working voltages for specific cross-isolation waveforms and insulation levels.

Table 5.

| UL | CSA | VDE | CQC |
| :---: | :---: | :---: | :---: |
| Recognized under UL 1577 Component Recognition Program ${ }^{1}$ | Approved under CSA Component Acceptance Notice 5A | Certified according to DIN V VDE V 0884-10 (VDE V 0884-10): 2006-12² | Approved under CQC11-471543-2015 |
| Single protection, 2500 V rms isolation voltage | Basic insulation per CSA 60950-1-03 and IEC 60950-1, 800 V rms ( 1131 V peak) maximum working voltage Reinforced insulation per CSA 60950-1-03 and IEC 60950-1, 400 V rms ( 566 V peak) maximum working voltage | Reinforced insulation, 560 V peak | Basic insulation per GB4943.1-2011, 760 V rms ( 1075 V peak) maximum working voltage, tropical climate, altitude $\leq 5000$ meters Reinforced insulation per GB4943.1-2011, 380 V rms ( 537 V peak) maximum working voltage, tropical climate, altitude $\leq 5000$ meters |
| File E214100 | File 205078 | File 2471900-4880-0001 | File CQC16001160843 |

${ }^{1}$ In accordance with UL1577, the ADuM3300/ADuM3301 are proof tested by applying an insulation test voltage $\geq 3000 \mathrm{~V}$ rms for 1 second (current leakage detection limit $=5 \mu \mathrm{~A}$ ).
${ }^{2}$ In accordance with DIN V VDE V 0884-10, the ADuM3300/ADuM3301 are proof tested by applying an insulation test voltage $\geq 1050 \mathrm{~V}$ peak for 1 second (partial discharge detection
limit $=5 \mathrm{pC})$. An asterisk ( ${ }^{*}$ ) marking branded on the component designates DIN V VDE V 0884-10 approval.

## INSULATION AND SAFETY-RELATED SPECIFICATIONS

Table 6.

| Parameter | Symbol | Value | Unit | Conditions |
| :--- | :--- | :--- | :--- | :--- |
| Rated Dielectric Insulation Voltage | L(I01) | 2500 | V rms | 1-minute duration |
| Minimum External Air Gap (Clearance) | mm | Measured from input terminals to output terminals, <br> shortest distance through air |  |  |
| Minimum External Tracking (Creepage) | 8.1 min | mm | Measured from input terminals to output terminals, <br> shortest distance path along body |  |
| Minimum Internal Gap (Internal Clearance) | CTI | 0.017 min | mm | Insulation distance through insulation <br> DIN IEC 112/VDE 0303 Part 1 |
| Tracking Resistance (Comparative Tracking Index) <br> Isolation Group | V | DIIa | Material Group (DIN VDE 0110, 1/89, Table 1) |  |

## DIN V VDE V 0884-10 (VDE V 0884-10) INSULATION CHARACTERISTICS

These isolators are suitable for reinforced electrical isolation only within the safety limit data. Maintenance of the safety data is ensured by protective circuits. The asterisk (*) marking on the package denotes DIN V VDE V 0884-10 approval for 560 V peak working voltage.

Table 7.

| Description | Conditions | Symbol | Characteristic | Unit |
| :---: | :---: | :---: | :---: | :---: |
| Installation Classification per DIN VDE 0110 |  |  |  |  |
| For Rated Mains Voltage $\leq 150 \mathrm{~V}$ rms |  |  | I to IV |  |
| For Rated Mains Voltage $\leq 300 \mathrm{~V}$ rms |  |  | I to III |  |
| For Rated Mains Voltage $\leq 400 \mathrm{~V}$ rms |  |  | I to II |  |
| Climatic Classification |  |  | 40/105/21 |  |
| Pollution Degree per DIN VDE 0110, Table 1 |  |  | 2 |  |
| Maximum Working Insulation Voltage |  | VIorm | 560 | $\checkmark$ peak |
| Input-to-Output Test Voltage, Method B1 | $V_{\text {IORM }} \times 1.875=V_{\text {PR, }}, 100 \%$ production test, $\mathrm{t}_{\mathrm{m}}=1 \mathrm{sec}$, partial discharge $<5 \mathrm{pC}$ | $V_{\text {PR }}$ | 1050 | $\checkmark$ peak |
| Input-to-Output Test Voltage, Method A | $\mathrm{V}_{\text {IORM }} \times 1.6=\mathrm{V}_{\text {PR, }}, \mathrm{t}_{\mathrm{m}}=60 \mathrm{sec}$, partial discharge $<5 \mathrm{pC}$ | $V_{\text {PR }}$ |  |  |
| After Environmental Tests Subgroup 1 |  |  | 896 | $\checkmark$ peak |
| After Input and/or Safety Test Subgroup 2 and Subgroup 3 | $\mathrm{V}_{\text {IORM }} \times 1.2=\mathrm{V}_{\text {PR, }}, \mathrm{t}_{\mathrm{m}}=60 \mathrm{sec}$, partial discharge $<5 \mathrm{pC}$ |  | 672 | $\checkmark$ peak |
| Highest Allowable Overvoltage | Transient overvoltage, $\mathrm{t}_{\text {TR }}=10$ seconds | $V_{\text {TR }}$ | 4000 | $\checkmark$ peak |
| Safety-Limiting Values | Maximum value allowed in the event of a failure (see Figure 3) |  |  |  |
| Case Temperature |  | Ts | 150 | ${ }^{\circ} \mathrm{C}$ |
| Side 1 Current |  | Is 1 | 265 | mA |
| Side 2 Current |  | Is2 | 335 | mA |
| Insulation Resistance at $\mathrm{T}_{5}$ | $\mathrm{V}_{10}=500 \mathrm{~V}$ | Rs | $>10^{9}$ | $\Omega$ |



Figure 3. Thermal Derating Curve, Dependence of Safety-Limiting Values with Case Temperature per DIN V VDE V 0884-10

RECOMMENDED OPERATING CONDITIONS
Table 8.

| Parameter | Symbol | Min | Max | Unit |
| :--- | :--- | :--- | :--- | :--- |
| Operating Temperature | $\mathrm{T}_{\mathrm{A}}$ | -40 | +105 | ${ }^{\circ} \mathrm{C}$ |
| Supply Voltages $^{1}$ | $\mathrm{~V}_{\mathrm{DD} 1}, \mathrm{~V}_{\mathrm{DD} 2}$ | 3.0 | 5.5 | V |
| Input Signal Rise and Fall Times |  |  | 1.0 | ms |

${ }^{1}$ All voltages are relative to their respective ground. See the DC Correctness and Magnetic Field Immunity section for information on immunity to external magnetic fields.

## ABSOLUTE MAXIMUM RATINGS

Ambient temperature $=25^{\circ} \mathrm{C}$, unless otherwise noted.
Table 9.

| Parameter | Symbol | Min | Max | Unit |
| :---: | :---: | :---: | :---: | :---: |
| Storage Temperature | $\mathrm{T}_{\text {St }}$ | -65 | +150 | ${ }^{\circ} \mathrm{C}$ |
| Ambient Operating Temperature | $\mathrm{T}_{\text {A }}$ | -40 | +105 | ${ }^{\circ} \mathrm{C}$ |
| Supply Voltages ${ }^{1}$ | $\mathrm{V}_{\mathrm{DD} 1}, \mathrm{~V}_{\mathrm{DD} 2}$ | -0.5 | +7.0 | V |
| Input Voltage ${ }^{1,2}$ | $\begin{aligned} & \mathrm{V}_{1 A}, V_{V_{B B},}, V_{1 C 1} \\ & V_{1 D}, V_{E 1}, V_{E 2} \end{aligned}$ | -0.5 | $\mathrm{V}_{\text {DII }}+0.5$ | V |
| Output Voltage ${ }^{1,2}$ | $\begin{aligned} & V_{O A}, V_{O B}, V_{O C}, \\ & V_{O D} \end{aligned}$ | -0.5 | $V_{\text {DDO }}+0.5$ | V |
| Average Output Current per Pin ${ }^{3}$ |  |  |  |  |
| Side 1 | lo1 | -23 | +23 | mA |
| Side 2 | lo2 | -30 | +30 | mA |
| Common-Mode Transients ${ }^{4}$ | $\mathrm{CM}_{\mathrm{H},} \mathrm{CM}_{\mathrm{L}}$ | -100 | +100 | kV/ $/ \mathrm{s}$ |

${ }^{1}$ All voltages are relative to their respective ground.
${ }^{2} V_{D D I}$ and $V_{D D O}$ refer to the supply voltages on the input and output sides of a given channel, respectively.
${ }^{3}$ See Figure 3 for maximum rated current values for various temperatures.
${ }^{4}$ Refers to common-mode transients across the insulation barrier. Commonmode transients exceeding the Absolute Maximum Rating can cause latch-up or permanent damage.

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

## ESD CAUTION



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

Table 10. Maximum Continuous Working Voltage ${ }^{1}$

| Parameter | Max | Unit | Constraint |
| :--- | :--- | :--- | :--- |
| AC Voltage, Bipolar Waveform | 565 | V peak | 50-year minimum lifetime |
| AC Voltage, Unipolar Waveform |  |  |  |
| $\quad$ Basic Insulation | 1131 | V peak | Maximum approved working voltage per IEC 60950-1 |
| $\quad$ Reinforced Insulation | 560 | V peak | Maximum approved working voltage per IEC 60950-1 and VDE V 0884-10 |
| DC Voltage |  |  |  |
| Basic Insulation | 1131 | V peak | Maximum approved working voltage per IEC 60950-1 |
| Reinforced Insulation | 560 | V peak | Maximum approved working voltage per IEC 60950-1 and VDE V 0884-10 |

${ }^{1}$ Refers to continuous voltage magnitude imposed across the isolation barrier. See the Insulation Lifetime section for more details.

Table 11. Truth Table (Positive Logic)

| VIX Input ${ }^{1}$ | VEx Input ${ }^{2}$ | V ${ }_{\text {dol }}$ State ${ }^{1}$ | V ${ }_{\text {dDo }}$ State ${ }^{1}$ | Vox Output ${ }^{1}$ | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: |
| H | H or NC | Powered | Powered | H |  |
| L | H or NC | Powered | Powered | L |  |
| $X$ | L | Powered | Powered | Z |  |
| X | H or NC | Unpowered | Powered | H | Outputs return to the input state within $1 \mu \mathrm{~s}$ of $\mathrm{V}_{\text {DDI }}$ power restoration |
| $X$ | L | Unpowered | Powered | Z |  |
| $X$ | X | Powered | Unpowered | Indeterminate | Outputs return to the input state within $1 \mu \mathrm{~s}$ of $\mathrm{V}_{\text {DDo }}$ power restoration if $\mathrm{V}_{\mathrm{EX}}$ state is H or NC Outputs return to high impedance state within 8 ns of $\mathrm{V}_{\text {DDO }}$ power restoration if $\mathrm{V}_{\mathrm{EX}}$ state is L |

[^2]
## PIN CONFIGURATIONS AND FUNCTION DESCRIPTIONS


*PIN 2 AND PIN 8 ARE INTERNALLY CONNECTED, AND CONNECTING BOTH TO GND ${ }_{1}$ IS RECOMMENDED.
**PIN 9 AND PIN 15 ARE INTERNALLY CONNECTED, AND CONNECTING
BOTH TO GND 2 IS RECOMMENDED. IN NOISY ENVIRONMENTS,
CONNECTING OUTPUT ENABLES (PIN 7 FOR ADuM3301 AND PIN 10 FOR
ALL MODELS) TO AN EXTERNAL LOGIC HIGH OR LOW IS RECOMMENDED.
Figure 4. ADuM3300 Pin Configuration
Table 12. ADuM3300 Pin Function Descriptions

| Pin No. | Mnemonic | Description |
| :---: | :---: | :---: |
| 1 | $\mathrm{V}_{\mathrm{DD} 1}$ | Supply Voltage for Isolator Side 1, 3.0 V to 5.5 V. |
| 2,8 | $\mathrm{GND}_{1}$ | Ground 1. Ground Reference for Isolator Side 1. |
| 3 | $V_{\text {IA }}$ | Logic Input A. |
| 4 | $V_{\text {IB }}$ | Logic Input B. |
| 5 | VIC | Logic Input C. |
| 6, 7, 11 | NC | No Connect. |
| 9,15 | $\mathrm{GND}_{2}$ | Ground 2. Ground Reference for Isolator Side 2. |
| 10 | $V_{\text {E2 }}$ | Output Enable 2. Active high logic input. $\mathrm{V}_{\mathrm{OA}}, \mathrm{V}_{\mathrm{OB}}$, and $\mathrm{V}_{\mathrm{OC}}$ outputs are enabled when $\mathrm{V}_{\mathrm{E} 2}$ is high or disconnected. $V_{\text {OA, }} V_{\text {Ob, }}$ and $V_{\text {Oc }}$ outputs are disabled when $V_{E 2}$ is low. In noisy environments, connecting $V^{E 2}$ to an external logic high or low is recommended. |
| 12 | Voc | Logic Output C. |
| 13 | Vob | Logic Output B. |
| 14 | VoA | Logic Output A. |
| 16 | $\mathrm{V}_{\mathrm{DD} 2}$ | Supply Voltage for Isolator Side 2, 3.0 V to 5.5 V. |


*PIN 2 AND PIN 8 ARE INTERNALLY CONNECTED,
AND CONNECTING BOTH TO GND 1 IS RECOMMENDED.
**PIN 9 AND PIN 15 ARE INTERNALLY CONNECTED,
AND CONNECTING BOTH TO GND 2 IS RECOMMENDED.
Figure 5. ADuM3301 Pin Configuration
Table 13. ADuM3301 Pin Function Descriptions

| Pin No. | Mnemonic | Description |
| :---: | :---: | :---: |
| 1 | VDD1 | Supply Voltage for Isolator Side 1, 3.0 V to 5.5 V. |
| 2, 8 | $\mathrm{GND}_{1}$ | Ground 1. Ground reference for Isolator Side 1. |
| 3 | $V_{\text {IA }}$ | Logic Input A. |
| 4 | $V_{\text {IB }}$ | Logic Input B. |
| 5 | Voc | Logic Output C. |
| 6, 11 | NC | No Connect. |
| 7 | $V_{E 1}$ | Output Enable 1. Active high logic input. Voc output is enabled when $V_{E 1}$ is high or disconnected. $V_{o c}$ is disabled when $\mathrm{V}_{\mathrm{E} 1}$ is low. In noisy environments, connecting $\mathrm{V}_{\mathrm{E} 1}$ to an external logic high or low is recommended. |
| 9, 15 | $\mathrm{GND}_{2}$ | Ground 2. Ground reference for Isolator Side 2. |
| 10 | $\mathrm{V}_{\mathrm{E} 2}$ | Output Enable 2. Active high logic input. $V_{\mathrm{OA}}$ and $\mathrm{V}_{\mathrm{OB}}$ outputs are enabled when $\mathrm{V}_{\mathrm{E} 2}$ is high or disconnected. $V_{\text {OA }}$ and $\mathrm{V}_{\text {OB }}$ outputs are disabled when $\mathrm{V}_{\mathrm{E} 2}$ is low. In noisy environments, connecting $\mathrm{V}_{\mathrm{E} 2}$ to an external logic high or low is recommended. |
| 12 | VIc | Logic Input C. |
| 13 | $V_{\text {ов }}$ | Logic Output B. |
| 14 | $V_{\text {OA }}$ | Logic Output A. |
| 16 | $\mathrm{V}_{\mathrm{DD} 2}$ | Supply Voltage for Isolator Side 1, 3.0 V to 5.5 V. |

## TYPICAL PERFORMANCE CHARACTERISTICS



Figure 6. Typical Input Supply Current per Channel vs. Data Rate (No Load)


Figure 7. Typical Output Supply Current per Channel vs. Data Rate (No Load)


Figure 8. Typical Output Supply Current per Channel vs. Data Rate (15 pF Output Load)


Figure 9. Typical ADuM3300 VDD1 Supply Current vs. Data Rate for 5 V and 3.3 V Operation


Figure 10. Typical ADuM3300 VDD2 Supply Current vs. Data Rate for 5 V and 3.3 V Operation


Figure 11. Typical ADuM3301 VDD1 Supply Current vs. Data Rate for 5 V and 3.3 V Operation


Figure 12. Typical ADuM3301 VDD2 Supply Current vs. Data Rate for 5 V and 3.3 V Operation


Figure 13. Propagation Delay vs. Temperature, C Grade

## APPLICATION INFORMATION

## PC BOARD LAYOUT

The ADuM3300/ADuM3301 digital isolator requires no external interface circuitry for the logic interfaces. Power supply bypassing is strongly recommended at the input and output supply pins (see Figure 14). Bypass capacitors are most conveniently connected between Pin 1 and Pin 2 for $V_{\text {DDI }}$ and between Pin 15 and Pin 16 for $V_{\text {DD2 }}$. The capacitor value should be between $0.01 \mu \mathrm{~F}$ and $0.1 \mu \mathrm{~F}$. The total lead length between both ends of the capacitor and the input power supply pin should not exceed 20 mm . Bypassing between Pin 1 and Pin 8 and between Pin 9 and Pin 16 should be considered unless the ground pair on each package side is connected close to the package.


Figure 14. Recommended Printed Circuit Board Layout
In applications involving high common-mode transients, care should be taken to ensure that board coupling across the isolation barrier is minimized. Furthermore, the board layout should be designed such that any coupling that does occur equally affects all pins on a given component side. Failure to ensure this could cause voltage differentials between pins exceeding the device's absolute maximum ratings, thereby leading to latch-up or permanent damage.
See the AN-1109 Application Note for board layout guidelines.

## SYSTEM-LEVEL ESD CONSIDERATIONS AND ENHANCEMENTS

System-level ESD reliability (for example, per IEC 61000-4-x) is highly dependent on system design, which varies widely by application. The ADuM3300/ADuM3301 incorporate many enhancements to make ESD reliability less dependent on system design. The enhancements include

- ESD protection cells added to all input/output interfaces.
- Key metal trace resistances reduced using wider geometry and paralleling of lines with vias.
- The SCR effect inherent in CMOS devices minimized by use of guarding and isolation technique between PMOS and NMOS devices.
- Areas of high electric field concentration eliminated using $45^{\circ}$ corners on metal traces.
- Supply pin overvoltage prevented with larger ESD clamps between each supply pin and its respective ground.

While the ADuM3300/ADuM3301 improve system-level ESD reliability, they are no substitute for a robust system-level design. See Application Note AN-793 ESD/Latch-Up Considerations with iCoupler Isolation Products for detailed recommendations on board layout and system-level design.

## PROPAGATION DELAY-RELATED PARAMETERS

Propagation delay is a parameter that describes the time it takes a logic signal to propagate through a component. The propagation delay to a logic low output can differ from the propagation delay to a logic high.


Figure 15. Propagation Delay Parameters
Pulse width distortion is the maximum difference between these two propagation delay values and is an indication of how accurately the input signal's timing is preserved.
Channel-to-channel matching refers to the maximum amount the propagation delay differs between channels within a single ADuM3300/ADuM3301 component.
Propagation delay skew refers to the maximum amount the propagation delay differs between multiple ADuM3300/ ADuM3301 components operating under the same conditions.

## DC CORRECTNESS AND MAGNETIC FIELD IMMUNITY

Positive and negative logic transitions at the isolator input cause narrow ( $\sim 1 \mathrm{~ns}$ ) pulses to be sent to the decoder via the transformer. The decoder is bistable and is, therefore, either set or reset by the pulses, indicating input logic transitions. In the absence of logic transitions at the input for more than $\sim 1 \mu \mathrm{~s}$, a periodic set of refresh pulses indicative of the correct input state is sent to ensure dc correctness at the output. If the decoder receives no internal pulses of more than about $5 \mu \mathrm{~s}$, the input side is assumed to be unpowered or nonfunctional, in which case the isolator output is forced to a default state (see Table 11) by the watchdog timer circuit.
The limitation on the ADuM3300/ADuM3301 magnetic field immunity is set by the condition in which induced voltage in the transformer's receiving coil is sufficiently large to either falsely set or reset the decoder. The following analysis defines the conditions under which this can occur. The 3 V operating condition of the ADuM3300/ADuM3301 are examined because it represents the most susceptible mode of operation.

The pulses at the transformer output have an amplitude greater than 1.0 V . The decoder has a sensing threshold at about 0.5 V , thus establishing a 0.5 V margin in which induced voltages can be tolerated. The voltage induced across the receiving coil is given by

$$
V=(-d \beta / d t) \sum \pi r_{n}^{2} ; n=1,2, \ldots, N
$$

where:
$\beta$ is magnetic flux density (gauss).
$r_{n}$ is the radius of the $\mathrm{n}^{\text {th }}$ turn in the receiving coil ( cm ).
$N$ is the number of turns in the receiving coil.
Given the geometry of the receiving coil in the ADuM330x and an imposed requirement that the induced voltage is at most $50 \%$ of the 0.5 V margin at the decoder, a maximum allowable magnetic field is calculated as shown in Figure 16.


Figure 16. Maximum Allowable External Magnetic Flux Density
For example, at a magnetic field frequency of 1 MHz , the maximum allowable magnetic field of 0.2 kgauss induces a voltage of 0.25 V at the receiving coil. This is about $50 \%$ of the sensing threshold and does not cause a faulty output transition. Similarly, if such an event were to occur during a transmitted pulse (and was of the worst-case polarity), it would reduce the received pulse from $>1.0 \mathrm{~V}$ to 0.75 V -still well above the 0.5 V sensing threshold of the decoder.
The preceding magnetic flux density values correspond to specific current magnitudes at given distances from the ADuM3300/ADuM3301 transformers. Figure 17 expresses these allowable current magnitudes as a function of frequency for selected distances. The ADuM3300/ADuM3301 are extremely immune and can be affected only by extremely large currents operated at high frequency very close to the component (see Figure 17). For the 1 MHz example noted, a 0.5 kA current would have to be placed 5 mm away from the ADuM3300/ ADuM3301 to affect the component's operation.


Figure 17. Maximum Allowable Current for Various Current-to-ADuM3300/ADuM3301 Spacings

Note that at combinations of strong magnetic field and high frequency, any loops formed by printed circuit board traces could induce error voltages sufficiently large enough to trigger the thresholds of succeeding circuitry. Care should be taken in the layout of such traces to avoid this possibility.

## POWER CONSUMPTION

The supply current at a given channel of the ADuM3300/ ADuM3301 isolator is a function of the supply voltage, the channel's data rate, and the channel's output load.
For each input channel, the supply current is given by

$$
\begin{array}{ll}
I_{D D I}=I_{D D I(Q)} & f \leq 0.5 f_{r} \\
I_{D D I}=I_{D D I(D)} \times\left(2 f-f_{r}\right)+I_{D D I}(Q) & f>0.5 f_{r}
\end{array}
$$

For each output channel, the supply current is given by

$$
\begin{array}{rl}
I_{D D O}=I_{D D O}(Q) & f \leq 0.5 f_{r} \\
I_{D D O}=\left(I_{D D O(D)}+\left(0.5 \times 10^{-3}\right) \times C_{L} \times V_{D D O}\right) \times\left(2 f-f_{r}\right)+I_{D D O(Q)} \\
f & f 0.5 f_{r}
\end{array}
$$

where:
$I_{D D I(D)}, I_{D D O}(\mathbb{D})$ are the input and output dynamic supply currents per channel ( $\mathrm{mA} / \mathrm{Mbps}$ ).
$C_{L}$ is the output load capacitance ( pF ).
$V_{D D O}$ is the output supply voltage $(\mathrm{V})$.
$f$ is the input logic signal frequency $(\mathrm{MHz})$; it is half of the input data rate expressed in units of Mbps .
$f_{r}$ is the input stage refresh rate (Mbps).
$I_{D D I(Q),} I_{D D O(Q)}$ are the specified input and output quiescent supply currents (mA).
To calculate the total $\mathrm{I}_{\mathrm{DD} 1}$ and $\mathrm{I}_{\mathrm{DD} 2}$ supply current, the supply currents for each input and output channel corresponding to $V_{D D 1}$ and $V_{D D 2}$ are calculated and totaled. Figure 6 provides perchannel input supply current as a function of data rate. Figure 7 and Figure 8 provide per-channel output supply current as a function of data rate for an unloaded output condition and for a 15 pF output condition, respectively. Figure 9 through Figure 12 provide total $V_{\text {DD1 }}$ and $V_{\text {DD2 }}$ supply current as a function of data rate for $\mathrm{ADuM} 3300 / \mathrm{ADuM} 3301$ channel configurations.

## INSULATION LIFETIME

All insulation structures eventually break down when subjected to voltage stress over a sufficiently long period. The rate of insulation degradation is dependent on the characteristics of the voltage waveform applied across the insulation. In addition to the testing performed by the regulatory agencies, Analog Devices executes an extensive set of evaluations to determine the lifetime of the insulation structure within the ADuM3300/ ADuM3301.

Analog Devices performs accelerated life testing using voltage levels higher than the rated continuous working voltage. Acceleration factors for several operating conditions are determined. These factors allow calculation of the time to failure at the actual working voltage.

The values shown in Table 10 summarize the peak voltage for 50 years of service life for a bipolar ac operating condition, and the maximum CSA/VDE approved working voltages. In many cases, the approved working voltage is higher than 50 -year service life voltage. Operation at these high working voltages can lead to shortened insulation life.
The insulation lifetime of the ADuM3300/ADuM3301 depends on the voltage waveform type imposed across the isolation barrier. The iCoupler insulation structure degrades at different rates depending on whether the waveform is bipolar ac, unipolar ac, or dc. Figure 18, Figure 19, and Figure 20 illustrate these different isolation voltage waveforms.
Bipolar ac voltage is the most stringent environment. The goal of a 50 -year operating lifetime under the ac bipolar condition determines the Analog Devices recommended maximum working voltage.

In the case of unipolar ac or dc voltage, the stress on the insulation is significantly lower. This allows operation at higher working voltages while still achieving a 50 -year service life. The working voltages listed in Table 10 can be applied while maintaining the 50 -year minimum lifetime, provided that the voltage conforms to either the unipolar ac or dc voltage cases. Any cross-insulation voltage waveform that does not conform to Figure 19 or Figure 20 should be treated as a bipolar ac waveform, and its peak voltage should be limited to the 50 -year lifetime voltage value listed in Table 10.

Note that the voltage presented in Figure 19 is shown as sinusoidal for illustration purposes only. It is meant to represent any voltage waveform varying between 0 V and some limiting value. The limiting value can be positive or negative, but the voltage cannot cross 0 V .


Figure 18. Bipolar AC Waveform

RATED PEAK VOLTAGE


Figure 19. Unipolar AC Waveform


Figure 20. DC Waveform

## OUTLINE DIMENSIONS



Figure 21. 16-Lead Standard Small Outline Package [SOIC_W] Wide Body (RW-16)
Dimensions shown in millimeters and (inches)

ORDERING GUIDE

| Model ${ }^{1,2}$ | Temperature Range ( ${ }^{\circ} \mathrm{C}$ ) | Number of Inputs, $V_{D D 1}$ Side | Number of Inputs, $V_{\text {DD } 2}$ Side | Maximum Data Rate (Mbps) | Maximum Propagation Delay, 5 V (ns) | Maximum <br> Pulse Width <br> Distortion (ns) | Package Option ${ }^{3}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADuM3300ARWZ | -40 to +105 | 3 | 0 | 1 | 100 | 40 | RW-16 |
| ADuM3300ARWZ-RL | -40 to +105 | 3 | 0 | 1 | 100 | 40 | RW-16 |
| ADuM3300BRWZ | -40 to +105 | 3 | 0 | 10 | 50 | 3 | RW-16 |
| ADuM3300BRWZ-RL | -40 to +105 | 3 | 0 | 10 | 50 | 3 | RW-16 |
| ADuM3300CRWZ | -40 to +105 | 3 | 0 | 90 | 32 | 2 | RW-16 |
| ADuM3300CRWZ-RL | -40 to +105 | 3 | 0 | 90 | 32 | 2 | RW-16 |
| ADuM3301ARWZ | -40 to +105 | 2 | 1 | 1 | 100 | 40 | RW-16 |
| ADuM3301ARWZ-RL | -40 to +105 | 2 | 1 | 1 | 100 | 40 | RW-16 |
| ADuM3301BRWZ | -40 to +105 | 2 | 1 | 10 | 50 | 3 | RW-16 |
| ADuM3301BRWZ-RL | -40 to +105 | 2 | 1 | 10 | 50 | 3 | RW-16 |
| ADuM3301CRWZ | -40 to +105 | 2 | 1 | 90 | 32 | 2 | RW-16 |
| ADuM3301CRWZ-RL | -40 to +105 | 2 | 1 | 90 | 32 | 2 | RW-16 |

${ }^{1} \mathrm{Z}=$ RoHS Compliant Part.
${ }^{2}$ Tape and reel are available. The addition of an "-RL" suffix designates a $13^{\prime \prime}$ ( 1,000 units) tape and reel option.
${ }^{3}$ RW-16 = 16-lead wide body SOIC.

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[^0]:    ${ }^{1}$ Protected by U.S. Patents 5,952,849; 6,873,065; 6,903,578; and 7,075,329.

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[^2]:    ${ }^{1} V_{I X}$ and $V_{O X}$ refer to the input and output signals of a given channel ( $\mathrm{A}, \mathrm{B}$, or C ). $\mathrm{V}_{E X}$ refers to the output enable signal on the same side as the $\mathrm{V}_{\mathrm{OX}}$ outputs. $\mathrm{V}_{D D I}$ and $\mathrm{V}_{D D O}$ refer to the supply voltages on the input and output sides of the given channel, respectively.
    ${ }^{2}$ In noisy environments, connecting $\mathrm{V}_{\mathrm{EX}}$ to an external logic high or low is recommended.

