## DESCRIPTION

This demonstration board allows the user to evaluate the LTC ${ }^{\circledR 1066-1 ~ s w i t c h e d-c a p a c i t o r ~ f i l t e r ~ o v e r ~ t h e ~ f u l l ~ o p-~}$ erational range. This board demonstrates proper layout,
bypassing, buffering and clock line routing to achieve best performance from the LTC filter products. Gerber files for this circuit board are available. Call the LTC factory.

## BOARD PHOTO



## PACKAGE DIAGRAM



## DEMO MANUAL DC048

## LTC1066-1 FILTER BOARD

## SCHEmATIC DIAGRAM



## PARTS LIST

| REFERENCE DESIGNATOR | QUANTITY | PART NUMBER | DESCRIPTION | VENDOR | PHONE NUMBER |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{C}_{\mathrm{F}, \mathrm{C}}$ ( (Note 1) | 2 | GRM42-6Y5V105Z025AL | Cap, 1.0uF, 25V, 10\% | Murata-Erie | (814) 237-1431 |
| C2, C3, C4, C5 | 4 | VJ1206Y104KXAMT | Cap, 0.1限, 50V, 10\% | Vitramon | (203) 268-6261 |
| D1, D2 | 2 | MBRS130LT3 | Schottky Diode | Motorola | (602) 244-3576 |
| E1 to E8 | 8 | 1502-2 | Terminal | Keystone | (718) 956-0666 |
| J1 | 1 | 227699-3 | PCB Mount BNC | AMP | (717) 564-0100 |
| JP1, JP2 | 2 | TSW-103-07-G-S | Header | Samtec | (812) 944-6744 |
| $\mathrm{R}_{\mathrm{F}}$ | 1 | 5043EM20K00J | Res, 20k, 1/4W, 5\% | Philips | (817) 325-7871 |
| $\mathrm{R}_{\text {IN }}$ | 1 | CD 1/4Z | Res, 0, 1/4W | SEI | (919) 850-9500 |
| R2 | 1 | CR32-201J | Res, 200 $2,1 / 8 \mathrm{~W}, 5 \%$ | AVX | (803) 448-9411 |
| R3, R4 | 2 | 5043EM4K990F | Res, 4.99k, 1/4W, 1\% | Philips | (817) 325-7871 |
| R5, R6 | 2 | CR32-200J | Res, $20 \Omega, 1 / 8 \mathrm{~W}, 5 \%$ | AVX | (803) 448-9411 |
| R7 | 1 | CR32-103J | Res, 10k, 1/8W, 5\% | AVX | (803) 448-9411 |
| U1 | 1 | LTC1066-1CS | IC | LTC | (408) 432-1900 |
|  | 2 | SNT-100-BK-T | Shunt | Samtec | (812) 944-6744 |

Note 1: For filter cutoff frequencies as low as 10 Hz , use a $\mathrm{C}_{\mathrm{F}}=33 \mu \mathrm{~F}$ nonpolarized surface mount electrolytic capacitor, Sanyo 16CV33NP (619) 661-6835 or equivalent.

# DEMO MANUAL DC048 <br> LTC1066-1 FILTER BOARD 

## DCO48 OPERATION

## DEMO BOARD OPERATION AND CONNECTION HINTS

When using the LTC1066-1 switched-capacitor filter evaluation board, the following steps should be followed to ensure correct operation. This demo board was designed for either single or dual supply operation.

## Step 1: Connecting Power Supply Lines

For dual supply operation connect $\mathrm{V}^{+}$supply to E3 and $\mathrm{V}^{-}$ supply to E8. The power supply ground is connected to E4 and JP2 shorts E7 to SGND.
For single supply operation connect $\mathrm{V}^{+}$supply to E3 and the power supply ground to E4. JP2 shorts E8 to SGND. The potential at FGND (E7) is $\mathrm{V}^{+} / 2$.
Note the $20 \Omega$ resistors (R5, R6) that isolate the filter supply lines from the buffer supply lines. This may effect the output swing slightly (the maximum output voltage swing will be reduced by $20 \Omega \times I_{\text {OUT }}$ ).

## Step 2: Clock Input

Any TTL or CMOS clock source with a square wave output and $50 \%$ duty cycle $( \pm 10 \%)$ is adequate to connect to J 1 (clock input). For single supply operation $>6 \mathrm{~V}$, the clock high level should be $>65 \%$ of $\mathrm{V}^{+}$. See Table 1 for more detail on high and low threshold values.
The $200 \Omega$ resistor (R2) between J 1 and pin 9 slows down the rise and fall times of the clock to further reduce charge coupling.

Table 1. Clock Source High and Low Threshold Levels

| POWER SUPPLY | HIGH LEVEL | LOW LEVEL |
| :--- | :---: | :---: |
| Dual Supply $= \pm 7.5 \mathrm{~V}$ | 2.18 V | 0.5 V |
| Dual Supply $= \pm 5 \mathrm{~V}$ | 1.45 V | 0.5 V |
| Dual Supply $= \pm 2.5 \mathrm{~V}$ | 0.73 V | -2.0 V |
| Single Supply $=12 \mathrm{~V}$ | 7.80 V | 6.5 V |
| Single Supply $=5 \mathrm{~V}$ | 1.45 V | 0.5 V |

## Step 3: Ratio 50:1/100:1

The DC level at pin 8 determines the ratio of the clock-tofilter cutoff frequency. When pin 8 is connected to $\mathrm{V}^{+}$the clock-to-cutoff frequency ratio ( $\mathrm{f}_{\text {CLK }} / \mathrm{f}_{\text {CUTOFF }}$ ) is $50: 1$ and the filter response is elliptic. The design of the internal switched-capacitor filter was optimized for a $50: 1$ operation.

When pin 8 is connected to ground (or $1 / 2$ supply for single supply operation), the $\mathrm{f}_{\text {CLK }} / \mathrm{f}_{\text {CUTOFF }}$ ratio is equal to $100: 1$ and the filter response is pseudo-linear phase. When JP1 is not used and pin 2 of $\mathrm{JP1} 1$ is connected to $\mathrm{V}^{-}$(or SGND for single supply), the filter response is transitional Butterworth elliptic and the $\mathrm{f}_{\text {CLK }} / f_{\text {Cutoff }}$ ratio is equal to $100: 1$ (please refer to the Typical Performance Characteristics in the LTC1066-1 data sheet).
When JP1 shorts R7 to $\mathrm{V}^{+}$the ratio is $50: 1$. If JP1 shorts R7 to SGND then the ratio is $100: 1$.

Since the ratio is mechanically switched by JP1, a 10k protection resistor is placed between pin 8 and the DC source.

## Step 4: Input Connection

The input of LTC1066-1 is E1 $\left(\mathrm{V}_{\text {IN }}\right)$ which is referenced to E2 (SGND).
For single supply operation the input must be in the linear range of the filter. For example, in a single 5 V operation the linear range is 1.4 V to 3.6 V referenced to E 2 (SGND).

## Step 5: Output Connection

The output of the LTC1066-1 is available atE5 ( $\mathrm{V}_{\text {OUT }}$ ) which is referenced to E6 (SGND).

## Step 6: Compensation

If compensation is needed R1 and C1 should be installed. Compensation is recommended for the following cases shown in Table 2.
Table 2. Instances Where an $\mathbf{R}_{\boldsymbol{c}}$ Compensation (15pF in Series with $30 \mathrm{k} \Omega$ Pins 11,13 ) is Recommended, $\mathrm{f}_{\text {CLK }} / \mathrm{f}_{\text {Cutoff }}=50: 1$

| $V_{S}=$ Single $5 \mathrm{~V}($ AGND $=2 \mathrm{~V})$ | $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ | $\mathrm{f}_{\text {CUTOFF }} \geq 28 \mathrm{kHz}$ |
| :--- | :--- | :--- |
|  | $\mathrm{T}_{\mathrm{A}}=70^{\circ} \mathrm{C}$ | $\mathrm{f}_{\text {CUTOFF }} \geq 24 \mathrm{kHz}$ |
| $\mathrm{V}_{\mathrm{S}}= \pm 5 \mathrm{~V}$ | $\mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ | $\mathrm{f}_{\text {CUTOFF }} \geq 60 \mathrm{kHz}$ |
|  | $\mathrm{T}_{\mathrm{A}}=70^{\circ} \mathrm{C}$ | $\mathrm{f}_{\text {CUTOFF }} \geq 50 \mathrm{kHz}$ |
| $\mathrm{V}_{\mathrm{S}}= \pm 7.5 \mathrm{~V}$ | $\mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ | $\mathrm{f}_{\text {CUTOFF }} \geq 70 \mathrm{kHz}$ |
|  | $\mathrm{T}_{\mathrm{A}}=70^{\circ} \mathrm{C}$ | $\mathrm{f}_{\text {CUTOFF }} \geq 60 \mathrm{kHz}$ |

## Step 7: $\mathbf{R}_{\text {IN }}, \mathbf{C}_{\text {IN }}$

If $R_{F}$ is greater than $20 k, R_{I N}$ must be changed from a short to a value equal to $\mathrm{R}_{\mathrm{F}} \mathrm{C}_{\text {IN }}$ must be $0.1 \mu \mathrm{~F}$.

Please refer to the AC performance section under the Applications Information section in the LTC1066-1 data sheet.

## PCB LAYOUT AOD FILIM



Component Side Silkscreen


Solder Side Silkscreen


Component Side


Solder Side


Component Side Solder Mask


Solder Side Solder Mask

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