DESCRIPTION

Demonstration circuit 1317A-B is isolated input to high current output **1/8th Brick footprint** converter featuring the LT[®]1952 switching controller with Active Reset circuit. The Active Reset circuit can improve the efficiency in wide input voltage applications. Also, the Active Reset allows the implementation of self-driven synchronous secondary rectifiers in some applications.

The DC1317A-B converts isolated wide 18V to 72V input to 5V output and provides over 30A of output current depending on cooling. When determining the cooling requirements the actual input voltage range and continuous maximum output current must be taken into account. The converter operates at 200kHz with the peak efficiency greater than 94%. The DC1317 can be easily modified to generate output voltages in the range from 0.6V to 48V. The output currents are limited by total output power of up to 150W.

The available versions of DC1317A are: DC1317A-A, 34-75Vin to 3.3V, 35A DC1317A-B, 18-72Vin to 5V, 25A DC1317A-C, 18-72Vin to 12V, 10A DC1317A-D, 18-72Vin to 24V, 5A DC1317A-E, 36-72Vin to 5V, 12A DC1317A-F, 9-36Vin to 3.3V, 20A DC1317A-G, 9-36Vin to 12V, 5A DC1317A-H, 9-36Vin to 48V, 1.5A

The DC1317 circuit features soft-start which prevents output voltage overshoot on startup or when recovering from overload condition.

IT1952

The DC1317 has precise over-current protection circuit that allows for continuous operation under short circuit conditions. The low power dissipation under short circuit conditions insures high reliability even during short circuits.

The LT1952 can be synchronized to an external clock of up to 400kHz. Please refer to LT1952 data sheet for design details and applications information.

Design files for this circuit board are available. Call the LTC factory.

LT is a trademark of Linear Technology Corporation

PARAMETER	CONDITION	VALUE
Minimum Input Voltage	IOUT = 0A to 25A	18V
Maximum Input Voltage	IOUT = 0A to 25A	72V
V _{OUT}	V _{IN} = 18V to 72V, I _{OUT} = 0A to 25A (30Amax)	5V ±3%
Typical Output Ripple VOUT	V _{IN} = 18V to 72V, I _{OUT} = 0A to 25A	50mV _{P-P}
Nominal Switching Frequency		200kHz

Table 1. Performance Summary

QUICK START PROCEDURE

Demonstration circuit 1317 is easy to set up to evaluate the performance of LT1952 circuit. **Refer to Figure 1 for proper measurement equipment setup** and follow the procedure below:

NOTE: When measuring the input or output voltage ripple, care must be taken to avoid a long ground lead on the oscilloscope probe. Measure the input or output voltage ripple by touching the probe tip directly across the Vin or Vout and GND terminals. See Figure 2. for proper scope probe technique.

- With power off, connect the input power supply to Vin and GND. Make sure that the input power supply has sufficient current rating at minimum input voltage for the required output load.
- 2. Turn on the power at the input.

NOTE: Make sure that the input voltage does not exceed 72V.

3. Check for the proper output voltage. Vout = 5V.

If there is no output, temporarily disconnect the load to make sure that the load is not set too high.

- 4. Once the proper output voltage is established, adjust the load within the operating range and observe the output voltage regulation, ripple voltage, efficiency and other parameters.
- 5. The DC1317 is equipped with an output capacitor CSYS (470uF) that approximates typical system rail capacitance. If system board already has capacitance of similar value CSYS can be removed.

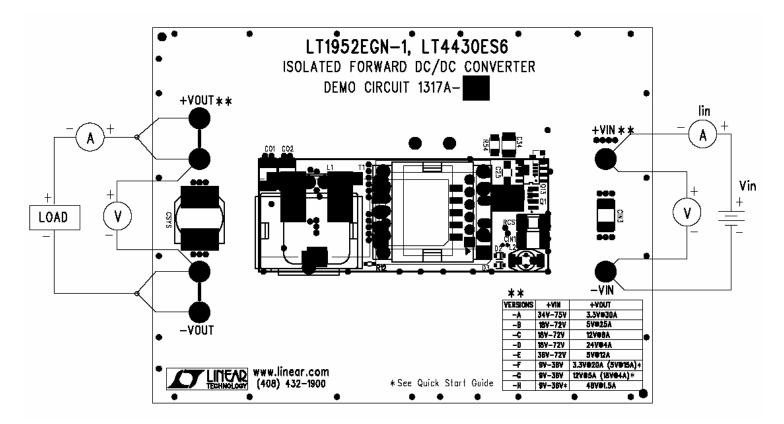


Figure 1. Proper Measurement Equipment Setup

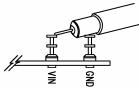


Figure 2. Scope Probe Placement for Measuring Input or Output Ripple

ACTIVE RESET CIRCUIT

The Active Reset circuit on DC1317A-B demo board consists of a small P-Channel MOSFET Q13 and reset capacitor C25. The MOSFET Q13 is used to connect the reset capacitor across the transformer T1 primary winding during the reset period when Q1 MOSFET is off. The voltage across capacitor C25 automatically adjusts with the duty cycle to provide complete transformer reset under all operating conditions.

Also the active reset circuit shapes the reset voltage into a square waveform that results in lower drain voltages for Q1 and Q2 MOSFETs. The lower MOSFET drain voltages allow lower voltage and lower Rdson MOSFETs to be used. The MOSFETs must be avalanche rated for the peak reset voltage. If non-avalanche rated MOSFETs are used a proper drain voltage derating should be used.

The main benefit of active reset circuit in the case of DC1317A-B demo board is high efficiency (shown in Figure 3), wide input range, high power density



and small size. To achieve such high efficiency all of the power components were carefully selected. Please consult LT factory for assistance if any changes to the circuit are required.

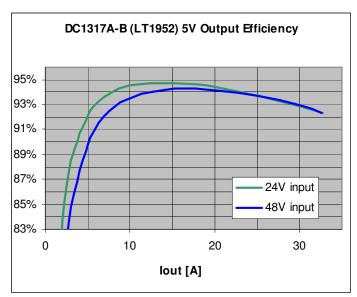


Figure 3. High efficiency of DC1317A-B allows the board to be used in thermally critical applications

OUTPUT LOAD STEP RESPONSE

The load step response of DC1317A-B, shown in Figure 4, is very fast even though relatively small amount of output capacitance is present (300uF ceramic and 470uF electrolytic). This is thanks to fast error amplifier of LT4430, optimal amount of current slope compensation of LT1952, fast opto coupler and fast error amplifier of LT1952. If higher load steps need to be handled more output capacitance can be added in order to keep the voltage transients at the desired level. Also, the loop response can be improved further by faster loop compensation.

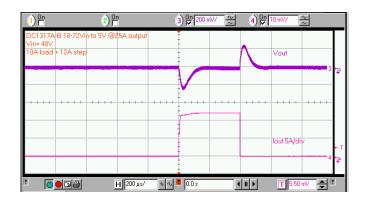


Figure 4. Fast transient response of DC1317 is superior to many competing power modules without the additional output capacitors.

SOFT START FUNCTION

The DC1317 features LT4430 opto coupler driver that has soft start function which produces monotonic startup ramp shown if Figure 5. The rise time of output voltage is controlled by capacitor C19 that is connected to OC (Overshoot Control) pin of LT4430. Also, the soft-start function will prevent input current surges even with full load at the output.

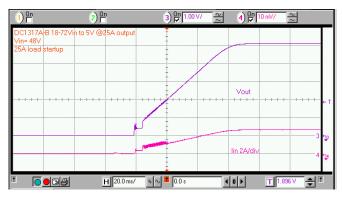


Figure 5. The LT4430 opto coupler driver produces monotonic output voltage rise at startup without output voltage overshoot.

DEBUGGING AND TESTING

The DC1317 can easily be tested and debugged by powering the bias circuit separately from the main power circuit. To place DC1317 into debug mode

remove the resistor R1 and connect 48V, 100mA power source to +Vb node (right side of R1). By doing this, the primary PWM controller LT1952 can be activated without the main primary power being applied to +Vin.

To activate the secondary side control circuit LT4430 diode OR a 10V, 100mA power source into collector of Q9 bias transistor.

Once the primary and secondary controllers are running the main power (+Vin) can be applied slowly while observing the switching waveforms and output voltage.

The input current supplying the power transformer T1 should not exceeded 100mA without the output load. If one of the MOSFETs is damaged, the input current will exceed 100mA.

PCB LAYOUT

The pcb layout should be carefully planned to avoid potential noise problems. The PCB layout for DC1317A can be used as a guide. Since demo board DC1317A has 8 versions the PCB layout has optional components that can be removed. Also, the PCB layout has a common schematic that is used just for the layout. The actual circuit schematic shows component values. The PCB layout schematic does not show the component values. In some cases, a different component like a diode is used in a place holder for a capacitor such as in the case of C6. Please modify the reference designators in your schematic to reflect the actual component used. The following simple PCB layout rules should be helpful.

If possible use solid ground planes on layers 2 and n-1. The ground planes will prevent the switching noise from coupling into sensitive lines.

Place sensitive lines on the inner layers that will be shielded by grounds on layers 2 and n-1.

Keep the loop formed by Q1, RCS1, Cin and T1 tight.

Keep the loop formed by Q2, Q3 and T1 tight.

Keep noise sensitive nodes like SD/VSEC, ROSC, FB, COMP, ISENSE, BLANK and DELAY as small as posible by placing the associated components close to the LT1952 and LT4430 chips.

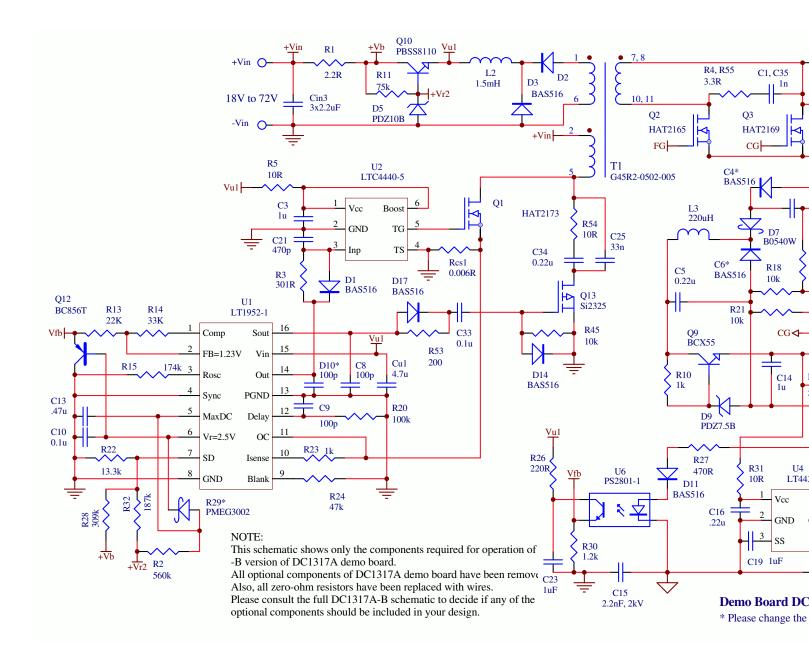
Use local vias for all components that connect to ground planes.

Do not place any traces on the layers 2 and n-1 to avoid ground planes from being compromised.

If the PCB layout has to be done on 2 or 4-layer PCB try to stay close to the guidelines outlined above. Also, maximize the ground connections between components by placing the components tight together.

Please contact LT factory for additional assistance.

QUICK START GUIDE FOR DEMONSTRATION CIRC ACTIVE RESET ISOLATED 18-72V INPUT TO 5V @25A DC/DC POV



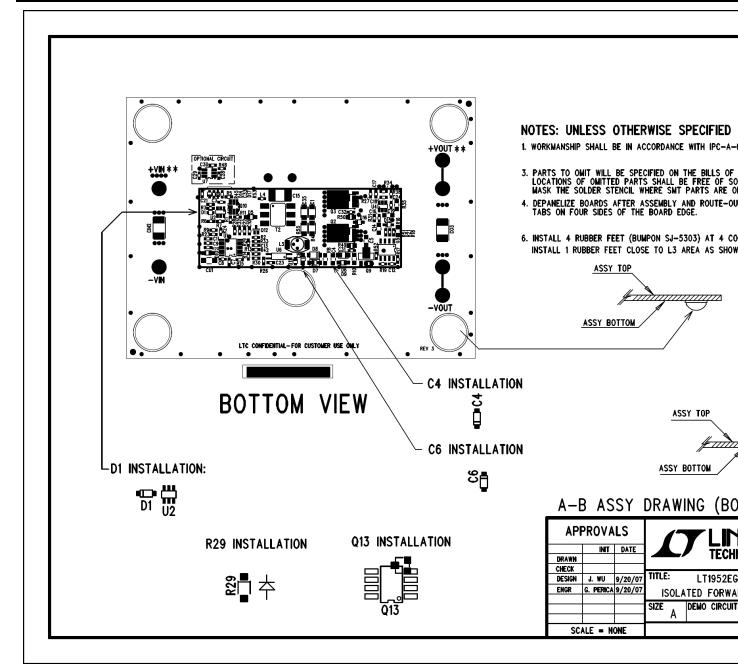


Q10 PBSS8110D L2 1.5mH DO1607B-155ML-C +VIN +VB R1 2.2 VU1 R4 R55 C1 C35 TP1 T1 G45R2-0502-005 CHAMPS TECH 3.3 1206 1nF 100V CIN1 CIN2 CIN3 2.2uF 100V TDK 1812 A1 ** D2 BAS516 SW1 +\ 7 VR2 L1 R11 75K K 1.90uH PA1393.202NL CG R49 opt. CO1 CO2 CO3 100uF 6.3V TDK 1812 SW2 BAS516 C31 opt. FG TP2 Ľ₽, +VIN Q3 HAT2169 1 _||<mark>opt.</mark> C32 11 VU1 R5 10 • D15 opt. 02 **B50** opt. SWO HAT2165 **D**16 U2 LTC4440ES6-5 R12 Q1 HAT2173 +VO1 C4 BAS516 VCOBOOST 6 C3 1uF VAUX B7 R8 1uF D6 opt. _1 Q9 BCX55 4 5 4 GND ΤG R9 0 D7 D8 opt. TS INF B0540W RCS1 0.006 KOA, TLR2H R10 1K C21 470pF L3 VFB 220uH R3 301 R6 OPT C7 220pF L4 10uH SW1 D1 C5 C6 0.22uF R16 33K 0805 T2 PE-68386 PULSE ENG. D9 PDZ10B R13 > 22K > BAS516 Q12 BC856T U1 LT1952EGN-1 ол R14 33K (BAS516 U5 LTC390 R18 10K R SOUT 1 16 Ŷ COMP SOUT R21 10K CS+ S SYNC VIN 15 R15 174K 2 C8 00p 2 5 R17 560 EB=1 23V -| VU1 D10 100pF CU1 4.7uF 14 CS- T ROSC OUT R42 13 SYNC PGND C9 100pF R20 100K 3 CG VU1 VAUX C10 0,1uF 5 SS/MAXDC DELAY 12 - C11 OPT R22 13.3K 6 11 4 vcc VR=2.5V oc R23_1K R26 220 7 SD/VSEC ISENSE 10 U6 PS2801-1 D11 BAS516 C23 1uF R27 470 R24 47K 8 GND BLANK 9 Ē ł VFB -|| 16V C13 0.47µF 10V R29 VAUX R31 3 U4 LT4430ES6 сĠ 10 R35 137K 1% **B33** 2 R30 2 1.2K 1 VCC OPTO +VO1 +VB PMEG3002AEB C16 0.22uF C17 15nF R34 3.4K C15 2200pF opt. R28 VR2 560K 5 GND COMP 250V 1812 309K C24 22pF R32 187K FB oc R53 200 D17 BAS516 C34 \uparrow Č19 1uF R38 18.2K 1% R54 * 10 1206 0.22uF 200V 1210 VERSION OPTIONAL FOOTPRINT FOR T SWO C25 33nF 200V 1206 000 - A C33 SW0 +VIN 0.1uF - B SW1 Q13 - C A1 7 NOTES: UNLESS OTHERWISE SPECIFIED - D SW2 1. ALL RESISTORS ARE 0603. - E R45 10K ALL CAPACITORS ARE 0603 3 9 10 - F OPTIONAL CIRCUIT * SEE QUICK START GUIDE. VU1 D14 BAS516 - G PA2720NL - H Ð U7 opt. C28 C29 opt. VCC BOOST 6 CONTRACT NO. LIN opt. CUSTOMER NOTICE 2 GND 5 TG LINEAR TECHNOLOGY HAS MADE A BEST EFFORT TO DESIGN A APPROVALS DATE TECH 3 INP CIRCUIT THAT MEETS CUSTOMER-SUPPLIED SPECIFICATIONS; HOWEVER, IT REMAINS THE CUSTOMER'S RESPONSIBILITY TO VERIFY PROPER AND RELIABLE OPERATION IN THE ACTUAL TS AWN 10/8/07 TILE 4 J. Wu LT1952EG ENGINEER C30 opt. G. Perica 10/8/07 **ISOLATED FORW** R48 opt. APPLICATION. COMPONENT SUBSTITUTION AND PRINTED CHECKED CIRCUIT BOARD LAYOUT MAY SIGNIFICANTLY AFFECT CIRCUIT SIZE CAGE CODE SOUT DWG PERFORMANCE OR RELIABILITY. CONTACT LINEAR TECHNOLOGY APPLICATIONS ENGINEERING FOR ASSISTANCE APPROVED THIS CIRCUIT IS PROPRIETARY TO LINEAR TECHNOLOGY AND Wednesday, February 25, 2009 SCALE: FILENAME: SUPPLIED FOR USE WITH LINEAR TECHNOLOGY PARTS.

QUICK START GUIDE FOR DEMONSTRATION CIRC ACTIVE RESET ISOLATED 18-72V INPUT TO 5V @25A DC/DC POV



QUICK START GUIDE FOR DEMONSTRATION CIRC ACTIVE RESET ISOLATED 18-72V INPUT TO 5V @25A DC/DC POV





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