

LTC3786 High Efficiency Synchronous Boost Converter

DESCRIPTION

Demonstration circuit DC1641 is DC/DC boost converter featuring the LTC3786 constant frequency current mode boost controller. The DC1641A operates over 5V to 24V input and provides over 5A at 6V input voltage and over 9A of output current at 10V input as shown in figure 3. The 350kHz constant frequency operation results in small and efficient circuit. The converter provides high output voltage accuracy (typically $\pm 3\%$) over wide load range with no minimum load requirement. The demonstration circuit can be easily modified to generate different output voltages. Please consult LTC factory for details.

The DC1641 has small circuit footprint. It is a high performance and cost effective solution for Telecom, Automotive and Power Over Ethernet applications.

Design files for this circuit board are available. Call the LTC factory.

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Table 1. Performance Summary

PARAMETER	CONDITION	VALUE
Minimum Input Voltage	IOUT = 0A to 5A	5V
Maximum Input Voltage	IOUT = 0A to 9A	24V
Vout	$V_{IN} = 5V$ to 24V, $I_{OUT} = 0A$ to 5A	24V ±3%
Typical Output Ripple VOUT	VIN = 12V, IOUT = 3A	100mVp_p
Nominal Switching Frequency		350kHz

QUICK START PROCEDURE

Demonstration circuit DC1641 is easy to set up to evaluate the performance of the LTC3786. For proper measurement equipment setup refer to Figure 1 and follow the procedure below:

NOTE: When measuring the input or output voltage ripple, care must be taken to minimize the length of oscilloscope probe ground lead. Measure the input or output voltage ripple by connecting the probe tip directly across the VIN or VOUT and GND terminals as shown in Figure 2.

- **1.** With power off, connect the input power supply to Vin and GND.
- **2.** Check the Vcc Bias voltage-setting jumper JP3. Move it into Vin position if required.

3. Turn the input power source on and slowly increase the input voltage. Be careful not to exceed 24V.

NOTE: Make sure that the input voltage Vin does not exceed 24V. If higher operating voltage is required, power components with higher voltage ratings should be used.

- **4.** Set the input voltage to 6V and check for the proper output voltage of 24V. If there is no output, temporarily disconnect the load to make sure that the load is not set too high.
- **5.** Once the proper output voltage is established, adjust the load and observe the output voltage regulation, ripple voltage, efficiency and other parameters.



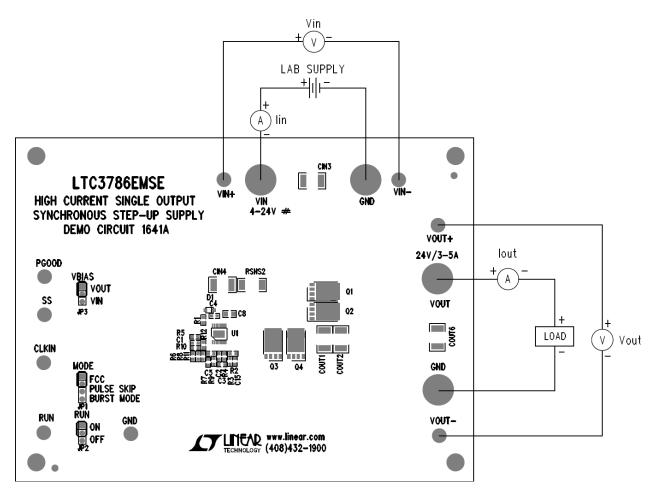


Figure 1. Proper Measurement Equipment Setup

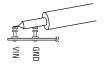


Figure 2. Measuring Input or Output Ripple

CHANGING THE OUTPUT VOLTAGE

To set the output voltage lower than 24V, change the bottom voltage divider resistors connected to LTC3786 FB pin (see the schematic on page 4). To get higher than 24V output voltage MOSFETs and capacitors with higher voltage ratings may be required. Please contact LTC factory for details. Note that the output voltage will start following the input voltage when input voltage is greater than the voltage set by the voltage divider. In this mode (wire mode) the synchronous output rectifier MOSFET is turned continuously ON. The only losses in this mode are due to the current sense resistor, inductor and output MOSFET DC resistance.



CONVERTER EFFICIENCY AND OUTPUT CURRENT

The DC1641 output current capability depends on the input voltage and proper cooling. Typical performance of DC1641A is shown in Figure 3. As can be seen from Figure 3, the output current capability depends on the input voltage. Also, when input voltage is equal to, or higher than the output voltage setting (wire mode) the efficiency is very high (99%) since the converter is not really running and the output MOSFET is shorting Vin to Vout.

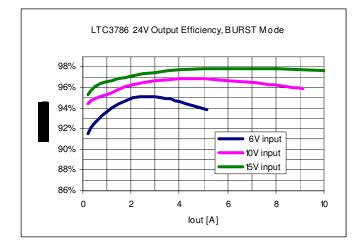


Figure 3. High efficiency of DC1641 allows the board to be used in thermally critical applications with outputs over 5A.

OUTPUT LOAD STEP RESPONSE

The load step response of DC1641A is very good even though relatively small amount of output capacitance is present at the output. If higher load steps need to be handled more output capacitance can be added in order to keep the voltage transients at the desired level. The load step transients are shown in Figure 4. Also, note that the load step response even from 0% load is excellent thanks to synchronous rectification.

INPUT VOLTAGE RANGE AND OUTPUT CURRENT

The minimum operating input voltage can be extended down to 4V or lower. However, the V_{GS} of MOSFETs and biasing must be considered.

The LTC3786 controller can be biased from $V_{\rm OUT}$ (move jumper JP3 to $V_{\rm OUT}$ position). The startup voltage in this case has to be higher than the $V_{\rm GS}$ threshold of MOSFETs. Also, the output current capability at minimum $V_{\rm IN}$ may be extended due to full bias of MOSFETs.

Keep in mind that the bias circuit power dissipation will be higher when LTC3786 is biased from V_{OUT} because V_{OUT} is higher than V_{IN} (except in WIRE mode). If the power dissipation in LTC3786 is too high an external NPN-Zener pre-regulator can also be used.

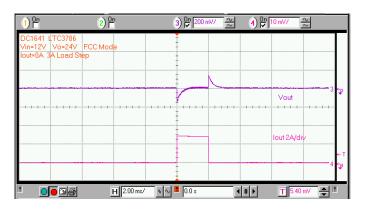


Figure 4. Fast transient response of DC1641 is achieved with a small amount of output capacitance.



SOFT START FUNCTION

The DC1641 features soft-start circuit that controls the inrush current and output voltage ramp at startup. The capacitor C_{15} (Figure 6) controls the startup period. The startup waveforms are shown in figure 5. Please note the output voltage step at about –7ms mark. The step is resulting from synchronous MOSFET being turned on. When the synchronous MOSFET is turned on the voltage drop across the synchronous MOSFET (body diode) is reduced from about 0.7V to lout*Rdson. While the synchronous MOSFET is in the off state (prior to –7ms mark) the synchronous MOSFET body diode is dissipating about 0.7V*lout or about 4.2W with 6A load. Please refer to the LTC3786 data sheet for details about proper handling of synchronous MOSFET in your application.

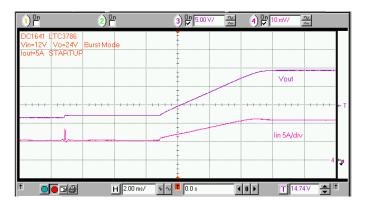


Figure 5. The DC1641 ramps the output slowly at startup without generating an input current surge.



J1 O VOUT 24V /3 - 5A 1530 McCarthy Blvd. 1620 McCarthy Blvd. Miprize, CA 95035 Miprize, CA 95035 Miprize, CA 95035 169454-0507 TTC Confidential-For Customer Use Only REV. APPROVED DATE GORAN P. 08-12-10 1 OF 1 4-24V # E2 VOUT-٨IN SHEET SYNCHRONOUS STEP-UP SUPPLY TITLE: SCHEMATIC HIGH CURRENT SINGLE OUTPUT VO. LTC3786EMSE DEMO CIRCUIT 1641A Wednesday, August 25, 2010 **REVISION HISTORY** SNScs cs RSNS1 0.005 DESCRIPTION CIN1 150uF 35V 35HVH150M PRODUCTION N⊦ VOUT RSNS2 0.005 +|(22uF 25V 1812 Ð 22uF 25V 1812 +⊳ 88 OPT +ECO REV 2 L1 3.3uH PA1494.362NL COUT1 COUT2 COUT3 COUT4 COUT6 IC NO. CIN2 CIN3 CIN4 CIN5 I DATE R7 OPT B11 OPT SIZE NVA ŝ 00T OPT 9 GORAN P. APPROVALS er a A.K SCALE = NONE PCB DES. APP ENG. BSC027N04 m LINEAR TECHNOLOGY HAS MADE A BET EFFORT TO BESIGN A CRUTI THAI MIERTS OF STORMER SHPERT A POINT AN A MERT A DISTORMER SHPERT AND REPRESE AND RELABLE OF STORMER SHPERT AND RETABLE OF STORMER SHPERT AND RETARLE OF STORMER AND RAITED PEPLLGATION. CORPORENT SHERT RETARLING MAD PANTEL CROAT BOARD LYOUT MY SCANFLAMITT, VAFFECT CRUAT PEPCORADOL OF AND RAILED RETARLING PARASISTARCE. 8 THIS CIRCUIT IS PROPRIETARY TO LINEAR TECHNOLOGY AND SUPPLIED FOR USE WITH LINEAR TECHNOLOGY PARTS. <u>9</u> BSC027N04 INTVCC **CUSTOMER NOTICE** 뛽숭 VBIAS C7 0.1uF D1 BAS140W 0 -D INTVCC C8 0805 0805 ₽₹ C4 0.1uF 0 ~~~ 100K Ð 16 15 4 ₽ 업 읃 SW TG BOOST VBIAS PGOOD INTVCC BG GND U1 LTC3786EMSE GND <u>_1</u> + -FOR VIN < VOUT ONLY, VOUT FOLLOWS VIN WHEN VIN >/= VOUT MODE/PLLIN VFB SENSE1+ SENSE1-FREQ RUN Ħ SS * * A - CONNECT TO RSNS1 KELVIN SENSE FOR SNS+/-**VOTE: UNLESS OTHERWISE SPECIFIED** ALL CAPACITORS ARE IN MICROFARADS, 0603. INSTALL SHUNTS AS SHOWN. R10 232K ¢ C1 OPT ₹I \rightarrow VBIAS C17 OPT R15 TBD 1. ALL RESISTORS ARE IN OHMS, 0603. INTVCC тоv C5 1000F то ч # - SURGE VOLTAGE UP TO 36V ی انک © 2 VBIAS Ч 99 ¥ 22 ¥ 0 18 NN ٨IN 4 62 우 SNScs+ Ś SS C19 1000pF Ð R19 38.3K R20 12.1K R4 16.9K C2 100pF ξ C15 0.1uF ٨N BURST MODE 10K 10K Ş 15nF C INTVCC 2 RUN F ъ В Ч -> FCC Ş 0 H2 GNDOE8 o g MODE RUN ss Es

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