# **ANALOG<br>DEVICES**

## **Micropower** Quad-Channel Digital Isolators

Data Sheet ADuM1440/ADuM1441/ADuM1442/ADuM1445/ADuM1446/ADuM1447

### **FEATURES**

- **Ultralow power operation**
- **3.3 V operation (typical)**
- **5.6 μA per channel quiescent current, refresh enabled 0.3 μA per channel quiescent current, refresh disabled 148 μA/Mbps per channel typical dynamic current 2.5 V operation (typical)** 
	- **3.1 μA per channel quiescent current, refresh enabled**
	- **0.1 μA per channel quiescent current, refresh disabled**
- **117 μA/Mbps per channel typical dynamic current Small, 16-lead QSOP and 20-Lead SSOP Bidirectional communication Up to 2 Mbps data rate (NRZ)**
- **High temperature operation: 125°C**
- **High common-mode transient immunity: >25 kV/μs**
- **Safety and regulatory approvals** 
	- **UL 1577 component recognition program**
	- **2500 V rms for 1 minute per UL 1577 QSOP package 3750V rms for 1 minute per UL 1577 SSOP package**
	- **CSA Component Acceptance Notice 5A**
	- **VDE certificate of conformity**
	- **DIN V VDE V 0884-10 (VDE V 0884-10):2006-12 VIORM = 565 VPEAK QSOP package**
	- **VIORM = 849 VPEAK SSOP package**
	- **IECEx and ATEX intrinsic safety**
	- **Sira 0518 II 1G Ex ia IIC Ga**

### **APPLICATIONS**

**General-purpose, low power multichannel isolation 1 MHz, low power peripheral interface (SPI) 4 mA to 20 mA loop process controls** 

### **GENERAL DESCRIPTION**

### The ADuM1440/ADuM1441/ADuM1442/ADuM1445/

ADuM1446/ADuM14471 are micropower, 4-channel digital isolators based on the Analog Devices, Inc., *i*Coupler® technology. Combining high speed, complementary metal oxide semiconductor (CMOS) and monolithic air core transformer technologies, these isolation components provide outstanding performance characteristics superior to the alternatives, such as optocoupler devices. As shown in Figure 3, in standard operating mode, when  $EN_x = 0$  (internal refresh enabled), the current per channel is less than 10 μA. When  $EN_x = 1$  (internal refresh disabled), the current per channel drops to less than 1 μA.

### The ADuM1440/ADuM1441/ADuM1442/ADuM1445/

ADuM1446/ADuM1447 family of quad 2.5 kV digital isolation devices are packaged in a small 16-lead QSOP and 20-lead SSOP,

1 Protected by U.S. Patents 5,952,849, 6,873,065, 7,075,329, 6,262,600. Other patents pending.

**Rev. E Document Feedback** 

**Information furnished by Analog Devices is believed to be accurate and reliable. However, no responsibility is assumed by Analog Devices for its use, nor for any infringements of patents or other rights of third parties that may result from its use. Specifications subject to change without notice. No license is granted by implication or otherwise under any patent or patent rights of Analog Devices. Trademarks and registered trademarks are the property of their respective owners.** 

### **FUNCTIONAL BLOCK DIAGRAMS**



freeing almost 70% of board space compared to isolators packages in wide body SOIC packages.

The devices withstand high isolation voltages and meet regulatory requirements, such as UL and CSA standards. In addition to the space savings, the ADuM1440/ADuM1441/ADuM1442/ ADuM1445/ADuM1446/ADuM1447 operate with supplies as low as 2.25 V.

Despite the low power consumption, all models of the ADuM1440/ ADuM1441/ADuM1442/ADuM1445/ADuM1446/ADuM1447 provide low, pulse width distortion at <8 ns. In addition, every model has an input glitch filter to protect against extraneous noise disturbances.



Figure 3. Typical Total Supply Current per Channel ( $V_{DDx} = 3.3 V$ )

**One Technology Way, P.O. Box 9106, Norwood, MA 02062-9106, U.S.A.** ©2013-2017 Analog Devices, Inc. All rights reserved. **Technical Support www.analog.com** 

### TABLE OF CONTENTS



### **REVISION HISTORY**



### **4/2015—Rev. C to Rev. D**



### **4/2015—Rev. B to Rev. C**



### **3/2015—Rev. A to Rev. B**





### **3/2014—Rev. 0 to Rev. A**



**10/2013—Revision 0: Initial Version** 

### **SPECIFICATIONS**

### **ELECTRICAL CHARACTERISTICS—3.3 V OPERATION**

All typical specifications are at  $T_A = 25^{\circ}C$ ,  $V_{DD1} = V_{DD2} = 3.3$  V. Minimum/maximum specifications apply over the entire recommended operating range of 3.0 V ≤ V<sub>DD1</sub> ≤ 3.6 V, 3.0 V ≤ V<sub>DD2</sub> ≤ 3.6 V, and  $-40^{\circ}$ C ≤ T<sub>A</sub> ≤ +125<sup>o</sup>C, unless otherwise noted. Switching specifications are tested with  $C_L = 15$  pF, and CMOS signal levels, unless otherwise noted.





<sup>1</sup> t<sub>PSK</sub> is the magnitude of the worst-case difference in t<sub>PHL</sub> and t<sub>PLH</sub> that is measured between units at the same operating temperature, supply voltages, and output load within the recommended operating conditions.



#### **Table 3. For All Models**



 $1 V_{DDx} = V_{DD1}$  or  $V_{DD2}$ .

 $^2$  |CM| is the maximum common-mode voltage slew rate that can be sustained while maintaining V<sub>ouT</sub> > 0.8 V<sub>DDx</sub>. The common-mode voltage slew rates apply to both rising and falling common-mode voltage edges.

### **ELECTRICAL CHARACTERISTICS—2.5 V OPERATION**

All typical specifications are at  $T_A = 25^{\circ}C$ ,  $V_{DD1} = V_{DD2} = 2.5$  V. Minimum/maximum specifications apply over the entire recommended operating range of 2.25 V ≤ V<sub>DD1</sub> ≤ 2.75 V, 2.25 V ≤ V<sub>DD2</sub> ≤ 2.75 V, and −40°C ≤ T<sub>A</sub> ≤ +125°C, unless otherwise noted. Switching specifications are tested with  $C_L = 15$  pF, and CMOS signal levels, unless otherwise noted.

#### **Table 4.**



 $1$  t<sub>PSK</sub> is the magnitude of the worst-case difference in t<sub>PHL</sub> or t<sub>PLH</sub> that is measured between units at the same operating temperature, supply voltages, and output load within the recommended operating conditions.



#### **Table 6. For All Models**



 $1 V_{DDx} = V_{DD1}$  or  $V_{DD2}$ .

 $^2$  |CM| is the maximum common-mode voltage slew rate that can be sustained while maintaining V<sub>ouT</sub> > 0.8 V<sub>DDx</sub>. The common-mode voltage slew rates apply to both rising and falling common-mode voltage edges.

**Table 7. Table 7.** 

### **ELECTRICAL CHARACTERISTICS—V**<sub>DD1</sub> = 3.3 V, V<sub>DD2</sub> = 2.5 V OPERATION

All typical specifications are at  $T_A = 25^{\circ}C$ ,  $V_{DD1} = 3.3$  V, and  $V_{DD2} = 2.5$  V. Minimum/maximum specifications apply over the entire recommended operating range of 3.0 V ≤ V<sub>DD1</sub> ≤ 3.6 V, 2.25 V ≤ V<sub>DD2</sub> ≤ 2.75 V, and  $-40^{\circ}$ C ≤ T<sub>A</sub> ≤ +125<sup>o</sup>C, unless otherwise noted. Switching specifications are tested with  $C_L = 15$  pF, and CMOS signal levels, unless otherwise noted.

For dc specifications and ac specifications, see Table 3 for Side 1 and see Table 6 for Side 2.



 $1$  t<sub>PSK</sub> is the magnitude of the worst-case difference in t<sub>PHL</sub> or t<sub>PLH</sub> that is measured between units at the same operating temperature, supply voltages, and output load within the recommended operating conditions.

**Table 8.** 

<b>Parameter</b>	Symbol	Min	Typ	Max	Unit	<b>Test Conditions/Comments</b>	
<b>SUPPLY CURRENT</b>						2 Mbps, no load	
ADuM1440/ADuM1445	I <sub>DD1</sub>		732	1000	μA	$EN_X = 0 V$ , $V_{IH} = V_{DD}$ , $V_{IL} = 0 V$	
	I <sub>DD2</sub>		337	750	μA	$EN_x = 0 V$ , $V_{1H} = V_{DD}$ , $V_{1L} = 0 V$	
ADuM1441/ADuM1446	I <sub>DD1</sub>		672	900	μA	$EN_X = 0 V$ , $V_{IH} = V_{DD}$ , $V_{IL} = 0 V$	
	I <sub>DD2</sub>		409	750	μA	$EN_x = 0 V$ , $V_{1H} = V_{DD}$ , $V_{1L} = 0 V$	
ADuM1442/ADuM1447	I <sub>DD1</sub>		612	900	μA	$EN_X = 0 V$ , $V_{IH} = V_{DD}$ , $V_{IL} = 0 V$	
	I <sub>DD2</sub>		480	750	μA	$EN_x = 0 V$ , $V_{1H} = V_{DD}$ , $V_{1L} = 0 V$	

### **ELECTRICAL CHARACTERISTICS—V**<sub>DD1</sub> = 2.5 V, V<sub>DD2</sub> = 3.3 V OPERATION

All typical specifications are at T<sub>A</sub> = 25°C, V<sub>DD1</sub> = 2.5, and V<sub>DD2</sub> = 3.3 V. Minimum/maximum specifications apply over the entire recommended operating range of 2.25 V ≤ V<sub>DD1</sub> ≤ 2.75 V, 3.0 V ≤ V<sub>DD2</sub> ≤ 3.6 V, and  $-40^{\circ}$ C ≤ T<sub>A</sub> ≤ +125 °C, unless otherwise noted. Switching specifications are tested with  $C_L = 15$  pF, and CMOS signal levels, unless otherwise noted.

For dc specifications and ac specifications, see Table 6 for Side 1 and see Table 3 for Side 2.





 $1$  t<sub>PSK</sub> is the magnitude of the worst-case difference in t<sub>PHL</sub> or t<sub>PLH</sub> that is measured between units at the same operating temperature, supply voltages, and output load within the recommended operating conditions.

#### **Table 10.**



### **PACKAGE CHARACTERISTICS**

#### **Table 11.**



<sup>1</sup> The device is considered a 2-terminal device: Pin 1 through Pin 8 are shorted together, and Pin 9 through Pin 16 are shorted together.

<sup>2</sup> Input capacitance is from any input data pin to ground.

### **REGULATORY INFORMATION**

See Table 20 and the Insulation Lifetime section for the recommended maximum working voltages for specific cross-isolation waveforms and insulation levels.

#### **Table 12. Safety Certifications**



<sup>1</sup> In accordance with UL 1577, each ADuM1440/ADuM1441/ADuM1442/ADuM1445/ADuM1446/ADuM1447 is proof tested by applying an insulation test voltage and measuring leakage during final production testing. QSOP package devices are tested at ≥3000 V rms for 1 sec with a current leakage detection limit = 5 μA. SSOP package devices are tested at  $\geq$ 4500 V rms for 1 sec with a current leakage detection limit = 10 µA.

<sup>2</sup> In accordance with DIN V VDE V 0884-10, each ADuM1440/ADuM1441/ADuM1442/ADuM1445/ADuM1446/ADuM1447 is proof tested by applying an insulation test voltage ≥1059 V<sub>PEAK</sub> for 1 second (partial discharge detection limit = 5 pC). The asterisk (\*) marked on the component designates DIN V VDE V 0884-10 approval.

### **INSULATION AND SAFETY RELATED SPECIFICATIONS**

#### **Table 13.**



### **DIN V VDE V 0884-10 (VDE V 0884-10):2006-12 INSULATION CHARACTERISTICS**

These isolators are suitable for reinforced electrical isolation within the safety limit data only. Maintenance of the safety data is ensured by protective circuits. The asterisk (\*) marked on packages denotes DIN V VDE V 0884-10 approval.

#### **Table 14. 16-Lead QSOP (RQ-16)**



#### **Table 15. 20-Lead SSOP (RS-20)**



### **INTRINSIC SAFETY**

The ADuM1440/ADuM1441/ADuM1442/ADuM1445/ADuM1446/ADuM1447 support intrinsic safety for IS to IS applications under IEC 60079-11, and carry ATEX and IECEx certifications. These devices do not currently support IS to non IS galvanic isolation applications due to the minimum insulation requirements of IEC60079-11.

### *Product Conformity Certificate*

Sira 16ATEX2265U and IECEx SIR 16.0091U

### *Special Conditions for Safe Use*

These components are certified to comply with IEC 60079-11:2011. When one of these components is used in equipment, the component is to be fitted on a PCB inside a suitable enclosure and recertified as equipment. The creepage and clearance distances across the isolating component have been evaluated, but the distances to other circuitry remain the responsibility of the user of the certified equipment.

This assembly is an isolating component between separate intrinsically safe circuits. It is recommended that the assembly be connected to suitably certified intrinsically safe circuits considering the entity parameters in Table 16.

#### **Table 16. IS Entity Parameters**



<sup>1</sup> L<sub>i</sub> is defined as input inductance, C<sub>i</sub> is input capacitance, P<sub>i</sub> is input power, U<sub>i</sub> is input voltage, and I<sub>i</sub> is input current.

The components (for example, digital isolators) being certified have the following safety ratings listed in Table 17. The temperature class is determined based on Table 17.

#### **Table 17. Temperature Class Information**







### **Recommended Operating Conditions**

#### **Table 18.**



<sup>1</sup> All voltages are relative to their respective grounds. See the DC Correctness section for information on immunity to external magnetic fields.

### ABSOLUTE MAXIMUM RATINGS

 $T_A = 25$ °C, unless otherwise noted.

#### **Table 19.**



<sup>1</sup> See Figure 4 for maximum safety power values for various temperatures. <sup>2</sup> Refers to common-mode transients across the insulation barrier. Common-mode transients exceeding the absolute maximum ratings can cause latch-up or permanent damage.

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

### **Table 21. Truth Table (Positive Logic) for all Models**

#### **Table 20. Maximum Continuous Working Voltage1**



<sup>1</sup> Refers to continuous voltage magnitude imposed across the isolation barrier. See the Insulation Lifetime section for more details.

#### **ESD CAUTION**



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.



 $1 H = high$ , L = low, X = don't care, and Z = high impedance.

<sup>&</sup>lt;sup>2</sup> V<sub>Ix</sub> and V<sub>Ox</sub> refer to the input and output signals of a given channel (A, B, C, or D).

 $3$  V<sub>DDI</sub> refers to the power supply on the input side of a given channel (A, B, C, or D).

 $4$  V<sub>DDO</sub> refers to the power supply on the output side of a given channel (A, B, C, or D).

<sup>&</sup>lt;sup>5</sup> Low input must follow a falling edge; otherwise, it can be in the default low state.

11845-104

ड 1845

### PIN CONFIGURATIONS AND FUNCTION DESCRIPTIONS



<sup>1</sup>PIN 2 AND PIN 8 ARE INTERNALLY CONNECTED. CONNECTING BOTH<br>TO GND<sub>1</sub> IS RECOMMENDED.<br><sup>2</sup>PIN 9 AND PIN 15 ARE INTERNALLY CONNECTED. CONNECTING<br>BOTH TO GND<sub>2</sub> IS RECOMMENDED.

*Figure 5. ADuM1440/ADuM1445QSOP Pin Configuration*



*Figure 6. ADuM1440/ADuM1445 SSOP Pin Configuration*

#### **Table 22. ADuM1440/ADuM1445 Pin Function Descriptions1**



<sup>1</sup> Reference the AN-1109 Application Note for specific layout guidelines.

 $2$  N/A = not applicable.







**CONNECTING BOTH TO GND<sub>1</sub> IS RECOMMENDED.**<br><sup>2</sup>PIN 11 AND PIN 19 ARE INTERNALLY CONNECTED.<br>CONNECTING BOTH TO GND<sub>2</sub> IS RECOMMENDED.

*Figure 8. ADuM1441/ADuM1446 SSOP Pin Configuration*



#### **Table 23. ADuM1441/ADuM1446 Pin Function Descriptions1**

<sup>1</sup> Reference the AN-1109 Application Note for specific layout guidelines.

 $2$  N/A = not applicable.



**1PIN 2 AND PIN 8 ARE INTERNALLY CONNECTED. CONNECTING BOTH** َعِ a GND<sub>1</sub> IS RECOMMENDED.<br>2PIN 9 AND PIN 15 ARE INTERNALLY CONNECTED. CONNECTING<br>BOTH TO GND<sub>2</sub> IS RECOMMENDED.

*Figure 9. ADuM1442/ADuM1447QSOP Pin Configuration*



11845-110  **CONNECTING BOTH TO GND1 IS RECOMMENDED. 2PIN 11 AND PIN 19 ARE INTERNALLY CONNECTED. CONNECTING BOTH TO GND2 IS RECOMMENDED.**

*Figure 10. ADuM1442/ADuM1447 SSOP Pin Configuration*



#### **Table 24. ADuM1442/ADuM1447 Pin Function Descriptions1**

<sup>1</sup> Reference the AN-1109 Application Note for specific layout guidelines.

 $2$  N/A = not applicable.

### TYPICAL PERFORMANCE CHARACTERISTICS





*Figure 12. Current Consumption per Output vs. Data Rate for 2.5 V, ENx = Low Operation*



*Figure 13. Current Consumption per Input vs. Data Rate for 3.3 V, ENx = Low Operation*



*Figure 14. Current Consumption per Output vs. Data Rate for 3.3 V, ENx = Low Operation*



*Figure 15. Current Consumption per Input vs. Data Rate for 2.5 V, ENx = High Operation*



*Figure 16. Current Consumption per Output vs. Data Rate for 2.5 V, ENx = High Operation*



*Figure 17. Current Consumption per Input vs. Data Rate for V<sub>DDX</sub> = 3.3 V*, *ENx = High Operation*



*Figure 18. Current Consumption per Output vs. Data Rate for V<sub>DDx</sub> = 3.3 V*, *ENx = High Operation*





*Figure 20. I<sub>DDx</sub> Current per Input vs. Data Input Voltage for V<sub>DDx</sub> = 2.5 V* 



*Figure 21. Typical Input and Output Supply Current per Channel vs. Temperature for V<sub>DDx</sub>* = 2.5 *V*, Data Rate = 100 *kbps* 







Figure 23. Typical Input and Output Supply Current per Channel vs. Temperature for  $V_{DDx} = 2.5$  V, Data Rate = 1000 kbps



Figure 24. Typical Input and Output Supply Current per Channel vs. Temperature for  $V_{DDx} = 3.3$  V, Data Rate = 1000 kbps



Figure 25. Typical Propagation Delay vs. Temperature for  $V_{DDx} = 3.3$  V or  $V_{DDx} = 2.5$  V



### APPLICATIONS INFORMATION **PCB LAYOUT**

#### The ADuM1440/ADuM1441/ADuM1442/ADuM1445/

ADuM1446/ADuM1447 digital isolators require no external interface circuitry for the logic interfaces. Power supply bypassing is strongly recommended at both input and output supply pins: V<sub>DD1</sub> and V<sub>DD2</sub> (see Figure 29). Choose a capacitor value between 0.01 µF and 0.1 µF. The total lead length between both ends of the capacitor and the input power supply pin must not exceed 20 mm.

Using proper PCB design choices, the ADuM1440/ADuM1441/ ADuM1442/ADuM1445/ADuM1446/ADuM1447 readily meets CISPR 22 Class A (and FCC Class A) emissions standards, as well as the more stringent CISPR 22 Class B (and FCC Class B) standards in an unshielded environment. Refer to the AN-1109 Application Note, *Recommendations for Control of Radiated Emissions with iCoupler Devices*, for PCB-related EMI mitigation techniques, including board layout and stack-up issues.



*Figure 30. Recommended Printed Circuit Board Layout, SSOP*

For applications involving high common-mode transients, it is important to minimize board coupling across the isolation barrier. Furthermore, design the board layout so that any coupling that does occur equally affects all pins on a given component side. Failure to ensure this can cause voltage differentials between pins exceeding the absolute maximum ratings of the device, thereby leading to latch-up or permanent damage.

### **PROPAGATION DELAY-RELATED PARAMETERS**

These products are optimized for minimum power consumption by eliminating as many internal bias currents as possible. As a result, the timing characteristics are more sensitive to operating voltage and temperature than in standard *i*Coupler products. Refer to Figure 21 through Figure 28 for the expected variation of these parameters.

Propagation delay is a parameter defined as the time it takes a logic signal to propagate through a component. The input-tooutput propagation delay time for a high-to-low transition can differ from the propagation delay time of a low-to-high transition.



Pulse width distortion is the maximum difference between these two propagation delay values and an indication of how accurately the timing of the input signal is preserved.

Channel-to-channel matching is the maximum amount of time the propagation delay differs between channels within a single ADuM1440/ADuM1441/ADuM1442/ADuM1445/ADuM1446/ ADuM1447 component.

Propagation delay skew is the maximum amount of time the propagation delay differs between multiple ADuM1440/ ADuM1441/ADuM1442/ADuM1445/ADuM1446/ADuM1447 components operating under the same conditions.

In edge-based systems, it is critical to reject pulses that are too short to be handled by the encode and decode circuits. The ADuM1440/ADuM1441/ADuM1442/ADuM1445/ADuM1446/ ADuM1447 implement a glitch filter to reject pulses less than the glitch filter operating threshold. This threshold depends on the operating voltage, as shown in Figure 26. Any pulse shorter than the glitch filter does not pass to the output. When the refresh circuit is enabled, pulses that match the glitch filter width have a small probability of being stretched until corrected by the next refresh cycle, or by the next valid data through that channel. To avoid issues with pulse stretching, observe the minimum pulse width requirements listed in the switching specifications.

### **DC CORRECTNESS** *Standard Operating Mode*

Positive and negative logic transitions at the isolator input cause narrow  $(-1 \text{ ns})$  pulses to be sent to the decoder using the transformer. The decoder is bistable and is, therefore, either set or reset by the pulses, indicating input logic transitions. When refresh and watchdog functions are enabled by pulling  $EN<sub>1</sub>$  and  $EN<sub>2</sub>$  low, in the absence of logic transitions at the input for more than ~140 µs, a periodic set of refresh pulses indicative of the correct input state is sent to ensure dc correctness at the output. If the decoder receives no internal pulses of more than approximately 200 µs, the input side is assumed unpowered or nonfunctional, in which case, the isolator watchdog circuit forces the output to a default state. The default state is either high as in the ADuM1440, ADuM1441, and ADuM1442 versions, or low as in the ADuM1445, ADuM1446, and ADuM1447 versions.

### **Low Power Operating Mode**

The ADuM1440/ADuM1441/ADuM1442/ADuM1445/

ADuM1446/ADuM1447 allow the refresh and watchdog functions to be disabled by pulling  $EN_1$  and  $EN_2$  to logic high for the lowest power consumption. These control pins must be set to the same value on each side of the component for proper operation.

In this mode, the current consumption of the chip drops to the microamp range. However, be careful when using this mode because dc correctness is no longer guaranteed at startup. For example, if the following sequence of events occurs:

- 1. Power is applied to Side 1
- 2. A high level is asserted on the V<sub>IA</sub> input
- 3. Power is applied to Side 2

The high on  $V_{IA}$  is not automatically transferred to the Side 2 VOA, and there can be a level mismatch that is not corrected until a transition occurs at VIA. After power is stable on each side and a transition occurs on the input of the channel, that channel's input and output state is correctly matched. This contingency can be addressed in several ways, such as sending dummy data, or toggling refresh on for a short period to force synchronization after turn on.

### **Recommended Input Voltage for Low Power Operation**

#### The ADuM1440/ADuM1441/ADuM1442/ADuM1445/

ADuM1446/ADuM1447 implement Schmitt trigger input buffers so that the devices operate cleanly in low data rate or noisy environments. Schmitt triggers allow a small amount of shoot through current when their input voltage is not approximate to either  $V_{DDx}$  or  $GND_x$  levels. This is because the two transistors are both slightly on when input voltages are in the middle of the supply range. For many digital devices, this leakage is not a large portion of the total supply current and may not be noticed; however, in the ultralow power ADuM1440/ADuM1441/ADuM1442/ ADuM1445/ADuM1446/ADuM1447, this leakage can be larger than the total operating current of the device and cannot be ignored.

To achieve optimum power consumption with the ADuM1440/ ADuM1441/ADuM1442/ADuM1445/ADuM1446/ ADuM1447, always drive the inputs as near to  $V_{DDx}$  or  $GND_x$  levels as possible. Figure 19 and Figure 20 illustrate the shoot through leakage of an input; therefore, whereas the logic thresholds of the input are standard CMOS levels, optimum power performance is achieved when the input logic levels are driven within 0.5 V of either V<sub>DDx</sub> or GND<sub>x</sub> levels.

### **MAGNETIC FIELD IMMUNITY**

The magnetic field immunity of the ADuM1440/ADuM1441/ ADuM1442/ADuM1445/ADuM1446/ADuM1447 is determined by the changing magnetic field, which induces a voltage in the receiving coil of the transformer large enough to either falsely set or reset the decoder. The following analysis defines the conditions under which this can occur. The 3.3 V operating condition of the ADuM1440/ADuM1441/ADuM1442/ ADuM1445/ADuM1446/ADuM1447 is examined because it represents the most typical mode of operation.

The pulses at the transformer output have an amplitude greater than 1.0 V. The decoder has a sensing threshold at about 0.5 V, thus establishing a 0.5 V margin in which induced voltages can be tolerated. The voltage induced across the receiving coil is given by

$$
V = (-d\beta/dt) \sum \pi r_n^2
$$
;  $n = 1, 2, ..., N$ 

where:

 $\beta$  is magnetic flux density (gauss).

 $r_n$  is the radius of the n<sup>th</sup> turn in the receiving coil (cm). *N* is the number of turns in the receiving coil.

Given the geometry of the receiving coil in the ADuM1440/ ADuM1441/ADuM1442/ADuM1445/ADuM1446/ADuM1447 and an imposed requirement that the induced voltage be, at most, 50% of the 0.5 V margin at the decoder, a maximum allowable magnetic field at a given frequency can be calculated. The result is shown in Figure 32.



Figure 32. Maximum Allowable External Magnetic Flux Density

For example, at a magnetic field frequency of 1 MHz, the maximum allowable magnetic field of 0.5 kgauss induces a voltage of 0.25 V at the receiving coil. This is about 50% of the sensing threshold and does not cause a faulty output transition. Similarly, if such an event occurred during a transmitted pulse (and was of the worst-case polarity), it would reduce the received pulse from >1.0 V to 0.75 V, still well above the 0.5 V sensing threshold of the decoder.

The preceding magnetic flux density values correspond to specific current magnitudes at given distances from the ADuM1440/ ADuM1441/ADuM1442/ADuM1445/ADuM1446/ADuM1447 transformers. Figure 33 shows these allowable current magnitudes as a function of frequency for selected distances. As shown, the ADuM1440/ADuM1441/ADuM1442/ADuM1445/ADuM1446/ ADuM1447 are extremely immune and can be affected only by extremely large currents operating at a high frequency very near to the component. For the 1 MHz example noted previously, a 1.2 kA current would have to be placed 5 mm away from the ADuM1440/ADuM1441/ADuM1442/ADuM1445/ADuM1446/ ADuM1447 to affect the operation of the component.



*Figure 33. Maximum Allowable Current for Various Current-to-ADuM1440/ ADuM1441/ADuM1442/ADuM1445/ADuM1446/ADuM1447 Spacings*

Note that at combinations of strong magnetic field and high frequency, any loops formed by PCB traces can induce error voltages sufficiently large enough to trigger the thresholds of succeeding circuitry. Take care in the layout of such traces to avoid this possibility.

### **POWER CONSUMPTION**

The supply current at a given channel of the ADuM1440/ ADuM1441/ADuM1442/ADuM1445/ADuM1446/ADuM1447 isolator is a function of the supply voltage, the data rate of the channel, and the output load of the channel.

For each input channel, the supply current is given by



For each output channel, the supply current is given by



 $I_{DDO} = (I_{DDO(D)} + (0.5 \times 10^{-3}) \times C_L \times V_{DDO}) \times (2f - f_r) + I_{DDO(Q)}$  $f > 0.5 f_r$ 

#### where:

*IDDI (D)*, *IDDO (D)* are the input and output dynamic supply currents per channel (mA/Mbps).

*IDDI (Q)*, *IDDO (Q)* are the specified input and output quiescent supply currents (mA).

*f* is the input logic signal frequency (MHz); it is half the input data rate, expressed in units of Mbps.

*fr* is the input stage refresh rate (Mbps).

*C<sub>L</sub>* is the output load capacitance (pF).

*V<sub>DDO</sub>* is the output supply voltage (V).

To calculate the total  $V_{DD1}$  and  $V_{DD2}$  supply current, the supply currents for each input and output channel corresponding to V<sub>DD1</sub> and V<sub>DD2</sub> are calculated and totaled. Figure 11 through Figure 18 show per channel supply currents as a function of data rate for an unloaded output condition.

The ADuM1440/ADuM1441/ADuM1442/ADuM1445/

ADuM1446/ADuM1447 devices are intended to operate at an ultralow current. This is achieved by operating the part at a low average data rate, either by bursting data at high speed at a low duty factor or by running low bit rates. If data is burst at high data rates, the part sits quiescent for the majority of the time, at low data rates, the power consumption approaches the quiescent power consumption. Table 25 shows the typical current for an input and output channel pair as well as the total power dissipated for that channel. The total power is summed across both sides of the device, so the power is being drawn from two different supplies. However, it shows how the power depends on the V<sub>DD</sub> values and the state of the refresh.

<b>State of</b>		<b>Typical Input</b> Channel	<b>Typical Output</b> Channel		
Refresh	V <sub>DDI</sub>	$I_{DDI(Q)}$	<b>V</b> <sub>DDO</sub> DDO(Q)		Power/Ch
Enabled	2.5V	$2.6 \mu A$	2.5V	$0.5 \mu A$	$7.8 \mu W$
	3.3V	$4.8 \mu A$	3.3V	$0.8 \mu A$	18.5 µW
<b>Disabled</b>	2.5V	$0.05 \mu A$	2.5V	$0.05 \mu A$	$0.3 \mu W$
	3.3V	$0.12 \mu A$	3.3V	$0.13 \mu A$	$0.8 \mu W$

**Table 25. Typical Total Power Dissipation Per Channel**

### **INSULATION LIFETIME**

All insulation structures eventually break down when subjected to voltage stress over a sufficiently long period. The rate of insulation degradation is dependent on the characteristics of the voltage waveform applied across the insulation. In addition to the testing performed by the regulatory agencies, Analog Devices carries out an extensive set of evaluations to determine the lifetime of the insulation structure within the ADuM1440/ ADuM1441/ADuM1442/ADuM1445/ADuM1446/ADuM1447.

Analog Devices performs accelerated life testing using voltage levels higher than the rated continuous working voltage. Acceleration factors for several operating conditions are determined. These factors allow calculation of the time to failure at the actual working voltage. The values shown in Table 20 summarize the peak voltage for 50 years of service life for a bipolar ac operating condition and the maximum CSA approved working voltages. In many cases, the approved working voltage is higher than the 50-year service life voltage. Operation at these high working voltages can lead to shortened insulation life, in some cases.

The insulation lifetime of the ADuM1440/ADuM1441/

ADuM1442/ADuM1445/ADuM1446/ADuM1447 depends on the voltage waveform type imposed across the isolation barrier. The *i*Coupler insulation structure degrades at different rates depending on whether the waveform is bipolar ac, unipolar ac, or dc. Figure 34, Figure 35, and Figure 36 illustrate these different isolation voltage waveforms.

Bipolar ac voltage is the most stringent environment. The goal of a 50-year operating lifetime under the ac bipolar condition determines the Analog Devices recommended maximum working voltage.

In the case of unipolar ac or dc voltage, the stress on the insulation is significantly lower. This allows operation at higher working voltages while still achieving a 50-year service life. The working voltages listed in Table 20 can be applied while maintaining the 50-year minimum lifetime provided the voltage conforms to either the unipolar ac or dc voltage case. Treat any cross-insulation voltage waveform that does not conform to Figure 35 or Figure 36 as a bipolar ac waveform, and limit its peak voltage to the 50-year lifetime voltage value listed in Table 20.

Note that the voltage presented in Figure 35 is shown as sinusoidal for illustration purposes only. It is meant to represent any voltage waveform varying between 0 V and some limiting value. The limiting value can be positive or negative, but the voltage cannot cross 0 V.



### OUTLINE DIMENSIONS



**COMPLIANT TO JEDEC STANDARDS MO-150-AE** *Figure 38. 20-Lead Shrink Small Outline Package [SSOP] (RS-20) Dimensions shown in millimeters*

**SEATING PLANE**

**8.20 7.80 7.40**

> **0.25 0.09**

**8° 4° 0°**

**5.30 5.00**

> **1.85 1.75 1.65**

**10**

**0.38**

ᅟᅭ

**1**

**0.65 BSC**

**COPLANARITY 0.65 BSC 0.22**<br>**0.10** 

<u> 1 | 1 | 1 | 1 | 1</u>

**0.05 MIN**

÷

**2.00 MAX**

**060106-A**

**0.95 0.75 0.55**

### **ORDERING GUIDE**



 $1 Z =$  RoHS Compliant Part.

 $^2$  Tape and reel is available. The addition of the –RL7 suffix indicates that the product is shipped on 7" tape and reel.



www.analog.com

Rev. E | Page 25 of 25

### **X-ON Electronics**

Largest Supplier of Electrical and Electronic Components

*Click to view similar products for* [Interface Development Tools](https://www.x-on.com.au/category/embedded-solutions/engineering-tools/analog-digital-ic-development-tools/interface-development-tools) *category:*

*Click to view products by* [Analog Devices](https://www.x-on.com.au/manufacturer/analogdevices) *manufacturer:* 

Other Similar products are found below :

[DP130SSEVM](https://www.x-on.com.au/mpn/texasinstruments/dp130ssevm) [ISO3086TEVM-436](https://www.x-on.com.au/mpn/texasinstruments/iso3086tevm436) [ADP5585CP-EVALZ](https://www.x-on.com.au/mpn/analogdevices/adp5585cpevalz) [CHA2066-99F](https://www.x-on.com.au/mpn/unitedmonolithicsemiconductors/cha206699f) [AS8650-DB](https://www.x-on.com.au/mpn/ams/as8650db) [MLX80104 TESTINTERFACE](https://www.x-on.com.au/mpn/melexis/mlx80104testinterface) [I2C-CPEV/NOPB](https://www.x-on.com.au/mpn/texasinstruments/i2ccpevnopb) [ISO35TEVM-434](https://www.x-on.com.au/mpn/texasinstruments/iso35tevm434) [416100120-3](https://www.x-on.com.au/mpn/digitalview/4161001203) [XR18910ILEVB](https://www.x-on.com.au/mpn/maxlinear/xr18910ilevb) [XR21B1421IL28-0A-EVB](https://www.x-on.com.au/mpn/maxlinear/xr21b1421il280aevb) [EVAL-ADM2491EEBZ](https://www.x-on.com.au/mpn/analogdevices/evaladm2491eebz) [MAXREFDES23DB#](https://www.x-on.com.au/mpn/maxim/maxrefdes23db) [MAX9286COAXEVKIT#](https://www.x-on.com.au/mpn/maxim/max9286coaxevkit) [MAX3100EVKIT](https://www.x-on.com.au/mpn/maxim/max3100evkit) [MAX13235EEVKIT](https://www.x-on.com.au/mpn/maxim/max13235eevkit) [MAX14970EVKIT#](https://www.x-on.com.au/mpn/maxim/max14970evkit) [XR21B1424IV64-0A-EVB](https://www.x-on.com.au/mpn/maxlinear/xr21b1424iv640aevb) [CMOD232+](https://www.x-on.com.au/mpn/maxim/cmod232) [MAX13042EEVKIT+](https://www.x-on.com.au/mpn/maxim/max13042eevkit) [MAX14838EVKIT#](https://www.x-on.com.au/mpn/maxim/max14838evkit) [MAXCAM705OV635AAA#](https://www.x-on.com.au/mpn/maxim/maxcam705ov635aaa) [MAX9205EVKIT](https://www.x-on.com.au/mpn/maxim/max9205evkit) [DS100BR111AEVK/NOPB](https://www.x-on.com.au/mpn/texasinstruments/ds100br111aevknopb) [DC241C](https://www.x-on.com.au/mpn/analogdevices/dc241c) [MAX9286RCARH3DB#](https://www.x-on.com.au/mpn/maxim/max9286rcarh3db) [MAX13035EEVKIT+](https://www.x-on.com.au/mpn/maxim/max13035eevkit) [DC1794A](https://www.x-on.com.au/mpn/analogdevices/dc1794a) [SN65HVS885EVM](https://www.x-on.com.au/mpn/texasinstruments/sn65hvs885evm) [EVB81112-A1](https://www.x-on.com.au/mpn/melexis/evb81112a1) [DFR0257](https://www.x-on.com.au/mpn/dfrobot/dfr0257) [ZLR964122L](https://www.x-on.com.au/mpn/microsemi/zlr964122l) [ZLR88822L](https://www.x-on.com.au/mpn/microsemi/zlr88822l) [DC196A-B](https://www.x-on.com.au/mpn/analogdevices/dc196ab) [DC196A-A](https://www.x-on.com.au/mpn/analogdevices/dc196aa) [DC327A](https://www.x-on.com.au/mpn/analogdevices/dc327a) [OM13585UL](https://www.x-on.com.au/mpn/nxp/om13585ul) [MAX16972AGEEVKIT#](https://www.x-on.com.au/mpn/maxim/max16972ageevkit) [MARS1-DEMO3-ADAPTER-GEVB](https://www.x-on.com.au/mpn/onsemiconductor/mars1demo3adaptergevb) [MAX7315EVKIT+](https://www.x-on.com.au/mpn/maxim/max7315evkit) [PIM511](https://www.x-on.com.au/mpn/pimoroni/pim511) [PIM536](https://www.x-on.com.au/mpn/pimoroni/pim536) [PIM517](https://www.x-on.com.au/mpn/pimoroni/pim517) [DEV-17512](https://www.x-on.com.au/mpn/sparkfun/dev17512) [STR-FUSB3307MPX-PPS-GEVK](https://www.x-on.com.au/mpn/onsemiconductor/strfusb3307mpxppsgevk) [MAXREFDES177#](https://www.x-on.com.au/mpn/maxim/maxrefdes177) [EVAL-ADM2567EEBZ](https://www.x-on.com.au/mpn/analogdevices/evaladm2567eebz) [EVAL-ADN4654EBZ](https://www.x-on.com.au/mpn/analogdevices/evaladn4654ebz) [MAX9275COAXEVKIT#](https://www.x-on.com.au/mpn/maxim/max9275coaxevkit) [MAX2202XEVKIT#](https://www.x-on.com.au/mpn/maxim/max2202xevkit)