

Low Noise, High Speed Precision Operational Amplifiers

FEATURES

- Guaranteed 4.5nV/√Hz 10Hz Noise
- Guaranteed 3.8nV/√Hz 1kHz Noise
- 0.1Hz to 10Hz Noise, 60nV_{P-P} Typical
- Guaranteed 7 Million Min Voltage Gain, R₁ = 2k
- Guaranteed 3 Million Min Voltage Gain, $R_1 = 600\Omega$
- Guaranteed 25µV Max Offset Voltage
- Guaranteed 0.6µV/°C Max Drift with Temperature
- Guaranteed 11V/µs Min Slew Rate (LT1037)
- Guaranteed 117dB Min CMRR

APPLICATIONS

- Low Noise Signal Processing
- Microvolt Accuracy Threshold Detection
- Strain Gauge Amplifiers
- Direct Coupled Audio Gain Stages
- Sine Wave Generators
- Tape Head Preamplifiers
- Microphone Preamplifiers

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DESCRIPTION

The LT®1007/LT1037 series features the lowest noise performance available to date for monolithic operational amplifiers: $2.5 \text{nV}/\sqrt{\text{Hz}}$ wideband noise (less than the noise of a 400Ω resistor), 1/f corner frequency of 2Hz and 60nV peak-to-peak 0.1Hz to 10Hz noise. Low noise is combined with outstanding precision and speed specifications: $10\mu\text{V}$ offset voltage, $0.2\mu\text{V}/^{\circ}\text{C}$ drift, 130dB common mode and power supply rejection, and 60MHz gain bandwidth product on the decompensated LT1037, which is stable for closed-loop gains of 5 or greater.

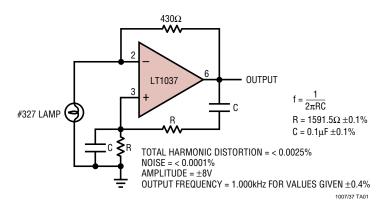
The voltage gain of the LT1007/LT1037 is an extremely high 20 million driving a $2k\Omega$ load and 12 million driving a 600Ω load to $\pm 10V$.

In the design, processing and testing of the device, particular attention has been paid to the optimization of the entire distribution of several key parameters. Consequently, the specifications of even the lowest cost grades (the LT1007C and the LT1037C) have been spectacularly improved compared to equivalent grades of competing amplifiers.

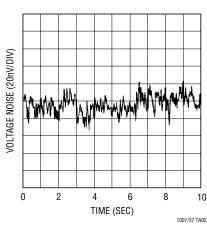
The sine wave generator application shown below utilizes the low noise and low distortion characteristics of the LT1037.

TYPICAL APPLICATION

Ultrapure 1kHz Sine Wave Generator



0.1Hz to 10Hz Noise

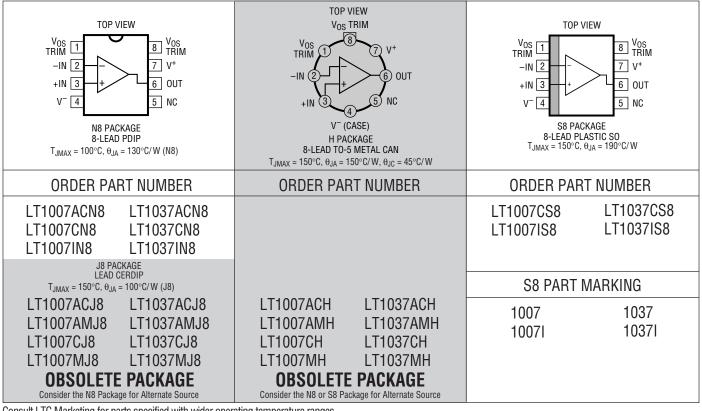


ABSOLUTE MAXIMUM RATINGS (Note 1)

Supply Voltage	±22V
Input Voltage	Equal to Supply Voltage
Output Short-Circuit Duration	Indefinite
Differential Input Current (Note	9) ±25mA
Storage Temperature Range	65°C to 150°C

Lead Temperature (Soldering, 10 sec.)......300°C Operating Temperature Range LT1007/LT1037AC, C 0°C to 70°C LT1007/LT1037I -40°C to 85°C LT1007/LT1037AM, M (OBSOLETE) -55°C to 125°C

PACKAGE/ORDER INFORMATION



Consult LTC Marketing for parts specified with wider operating temperature ranges.

ELECTRICAL CHARACTERISTICS $V_S = \pm 15V$, $T_A = 25^{\circ}C$, unless otherwise noted.

				1007AC/ 1037AC/		L1 L1			
SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
V _{0S}	Input Offset Voltage	(Note 2)		10	25		20	60	μV
$\frac{\Delta V_{0S}}{\Delta Time}$	Long Term Input Offset Voltage Stability	(Notes 3, 4)		0.2	1.0		0.2	1.0	μV/Mo
I _{OS}	Input Offset Current			7	30		12	50	nA
I _B	Input Bias Current			±10	±35		±15	±55	nA
e _n	Input Noise Voltage	0.1Hz to 10Hz (Notes 4, 6)		0.06	0.13		0.06	0.13	μV _{P-P}
	Input Noise Voltage Density	f ₀ = 10Hz (Notes 4, 5) f ₀ = 1000Hz (Note 4)		2.8 2.5	4.5 3.8		2.8 2.5	4.5 3.8	nV/√Hz nV/√Hz
i _n	Input Noise Current Density	$f_0 = 10$ Hz (Notes 4, 7) $f_0 = 1000$ Hz (Notes 4, 7)		1.5 0.4	4.0 0.6		1.5 0.4	4.0 0.6	pA/√Hz pA/√Hz



ELECTRICAL CHARACTERISTICS $V_S = \pm 15 V, \, T_A = 25^{\circ} C, \, unless \, otherwise \, noted.$

SYMBOL	PARAMETER	CONDITIONS		1007AC/ 11037AC/ TYP			T1007C/I, T1037C/I, TYP		UNITS
OTHIDOL	Input Resistance, Common Mode		IVIII		IIIAA	IVIIIV	5	ШЛЛ	GΩ
	Input Voltage Range		±11.0	±12.5		±11.0	±12.5		V
CMRR	Common Mode Rejection Ratio	V _{CM} = ±11V	117	130		110	126		dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 4V \text{ to } \pm 18V$	110	130		106	126		dB
A _{VOL}	Large-Signal Voltage Gain	$R_L \ge 2k$, $V_0 = \pm 12V$ $R_L \ge 1k$, $V_0 = \pm 10V$ $R_L \ge 600\Omega$, $V_0 = \pm 10V$	7.0 5.0 3.0	20.0 16.0 12.0		5.0 3.5 2.0	20.0 16.0 12.0		V/μV V/μV V/μV
V _{OUT}	Maximum Output Voltage Swing	$\begin{aligned} R_L &\geq 2k \\ R_L &\geq 600\Omega \end{aligned}$	±13.0 ±11.0	±13.8 ±12.5		±12.5 ±10.5	±13.5 ±12.5		V
SR	Slew Rate LT1007 LT1037	$R_L \ge 2k$ $A_{VCL} \ge 5$	1.7 11	2.5 15		1.7 11	2.5 15		V/µs V/µs
GBW	Gain Bandwidth LT1007 Product LT1037	$f_0 = 100 \text{kHz (Note 8)}$ $f_0 = 10 \text{kHz (Note 8) (A_{VCL} \ge 5)}$	5.0 45	8.0 60		5.0 45	8.0 60		MHz MHz
Z ₀	Open-Loop Output Resistance	$V_0 = 0V, I_0 = 0$		70			70		Ω
P _D	Power Dissipation LT1007 LT1037			80 80	120 130		80 85	140 140	mW mW

The ullet denotes the specifications which apply over the temperature range 0°C \leq T_A \leq 70°C, V_S = \pm 15V, unless otherwise noted.

			LT1007AC LT1037AC							
SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	MIN	TYP	MAX	UNITS
V _{OS}	Input Offset Voltage	(Note 2)	•		20	50		35	110	μV
$\Delta V_{OS} \over \Delta Temp$	Average Input Offset Drift	(Note 10)	•		0.2	0.6		0.3	1.0	μV/°C
I _{OS}	Input Offset Current		•		10	40		15	70	nA
I _B	Input Bias Current		•		±14	±45		±20	±75	nA
	Input Voltage Range		•	±10.5	±11.8		±10.5	±11.8		V
CMRR	Common Mode Rejection Ratio	$V_{CM} = \pm 10.5V$	•	114	126		106	120		dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 4.5 \text{V to } \pm 18 \text{V}$	•	106	126		102	120		dB
A _{VOL}	Large-Signal Voltage Gain	$R_L \ge 2k, V_0 = \pm 10V$ $R_L \ge 1k, V_0 = \pm 10V$	•	4.0 2.5	18.0 14.0		2.5 2.0	18.0 14.0		V/μV V/μV
V _{OUT}	Maximum Output Voltage Swing	$R_L \ge 2k$	•	±12.5	±13.6		±12.0	±13.6		V
P_{D}	Power Dissipation		•		90	144		90	160	mW

ELECTRICAL CHARACTERISTICS

The ullet denotes the specifications which apply over the temperature range $-40^{\circ}\text{C} \le T_{A} \le 85^{\circ}\text{C}$, $V_{S} = \pm 15V$, unless otherwise noted.

		LT1					
SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V _{OS}	Input Offset Voltage	(Note 2)	•		40	125	μV
$\frac{\Delta V_{OS}}{\Delta Temp}$	Average Input Offset Drift	(Note 10)	•		0.3	1.0	μV/°C
I _{OS}	Input Offset Current		•		20	80	nA
I _B	Input Bias Current		•		±25	±90	nA
	Input Voltage Range		•	±10	±11.7		V
CMRR	Common Mode Rejection Ratio	$V_{CM} = \pm 10.5V$	•	105	120		dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 4.5 V \text{ to } \pm 18 V$	•	101	120		dB
A _{VOL}	Large-Signal Voltage Gain	$R_L \ge 2k$, $V_0 = \pm 10V$ $R_L \ge 1k$, $V_0 = \pm 10V$	•	2.0 1.5	15.0 12.0		V/μV V/μV
V _{OUT}	Maximum Output Voltage Swing	$R_L \ge 2k$	•	±12.0	±13.6		V
P_{D}	Power Dissipation		•		95	165	mW

The ullet denotes the specifications which apply over the temperature range $-55^{\circ}\text{C} \le T_{A} \le 125^{\circ}\text{C}$, $V_{S} = \pm 15\text{V}$, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		LT1007AM/LT1037AM MIN TYP MAX		LT10	07M/LT1	037M Max	UNITS	
		+		IVIIIV			IVIIIV			
V_{OS}	Input Offset Voltage	(Note 2)	•		25	60		50	160	μV
ΔV_{OS}	Average Input Offset Drift	(Note 10)	•		0.2	0.6		0.3	1.0	μV/°C
ΔTemp										
I _{OS}	Input Offset Current		•		15	50		20	85	nA
I _B	Input Bias Current		•		±20	±60		±35	±95	nA
	Input Voltage Range		•	±10.3	±11.5		±10.3	±11.5		V
CMRR	Common Mode Rejection Ratio	$V_{CM} = \pm 10.3V$	•	112	126		104	120		dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 4.5 V \text{ to } \pm 18 V$	•	104	126		100	120		dB
A _{VOL}	Large-Signal Voltage Gain	$R_1 \ge 2k, V_0 = \pm 10V$		3.0	14.0		2.0	14.0		V/µV
VOL		$R_L \ge 1k, V_0 = \pm 10V$	•	2.0	10.0		1.5	10.0		V/µV
V _{OUT}	Maximum Output Voltage Swing	$R_L \ge 2k$	•	±12.5	±13.5		±12.0	±13.5		V
P_{D}	Power Dissipation		•		100	150		100	170	mW

For MIL-STD components, please refer to LTC 883C data sheet for test listing and parameters.

Note 1: Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

Note 2: Input Offset Voltage measurements are performed by automatic test equipment approximately 0.5 seconds after application of power. AM and AC grades are guaranteed fully warmed up.

Note 3: Long Term Input Offset Voltage Stability refers to the average trend line of Offset Voltage vs Time over extended periods after the first 30 days of operation. Excluding the initial hour of operation, changes in V_{OS} during the first 30 days are typically 2.5 μ V. Refer to typical performance curve.

Note 4: This parameter is tested on a sample basis only.

Note 5: 10Hz noise voltage density is sample tested on every lot. Devices 100% tested at 10Hz are available on request.

Note 6: See the test circuit and frequency response curve for 0.1Hz to 10Hz tester in the Applications Information section.

Note 7: See the test circuit for current noise measurement in the Applications Information section.

Note 8: This parameter is guaranteed by design and is not tested.

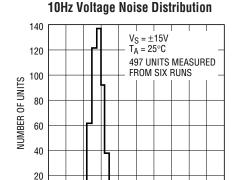
Note 9: The inputs are protected by back-to-back diodes. Current limiting resistors are not used in order to achieve low noise. If differential input voltage exceeds ± 0.7 V, the input current should be limited to 25mA.

Note 10: The Average Input Offset Drift performance is within the specifications unnulled or when nulled with a pot having a range of $8k\Omega$ to $20k\Omega$.

LINEAR

10

1007/37 G01

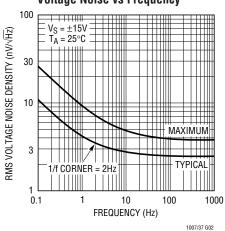


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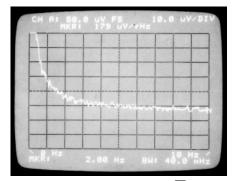
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Voltage Noise vs Frequency



0.02Hz to 10Hz RMS Noise. Gain = 50,000 (Measured on HP3582 Spectrum Analyzer)



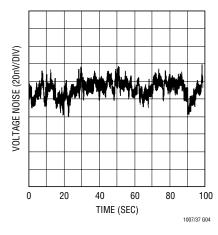
MARKER AT 2Hz (= 1/f CORNER) = $\frac{179 \mu V / \sqrt{\text{Hz}}}{50,000}$ = $3.59 \frac{\text{nV}}{\sqrt{\text{Hz}}}$

0.01Hz to 1Hz Peak-to-Peak Noise

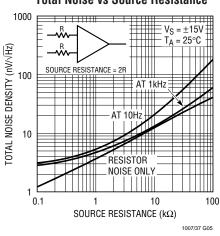
6 7 8 9

5

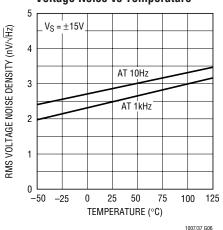
VOLTAGE NOISE DENSITY (nV/\sqrt{Hz})



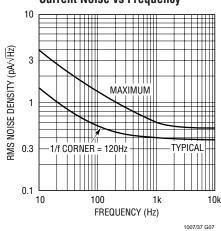
Total Noise vs Source Resistance



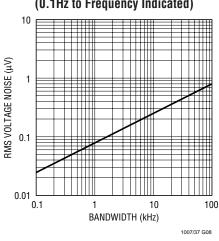
Voltage Noise vs Temperature



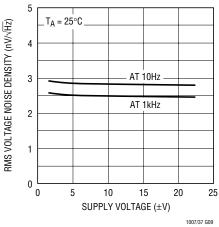
Current Noise vs Frequency



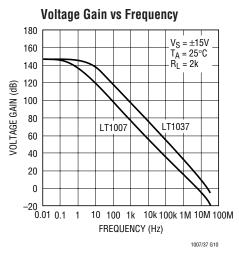
Wideband Voltage Noise (0.1Hz to Frequency Indicated)

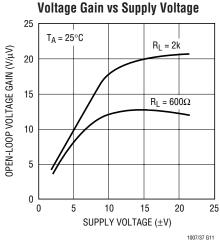


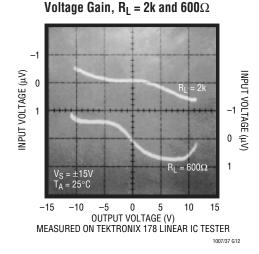
Voltage Noise vs Supply Voltage

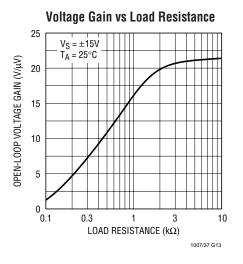


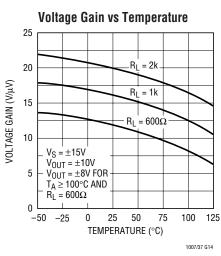


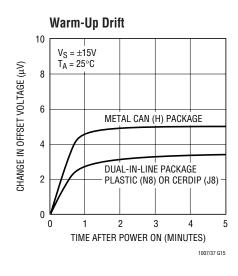


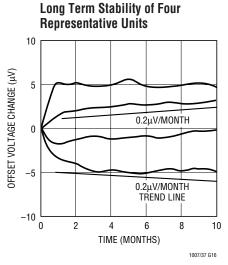


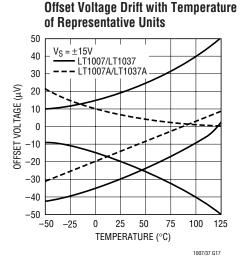


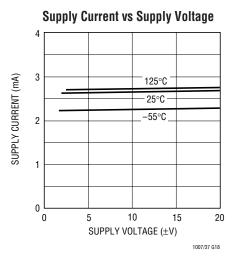








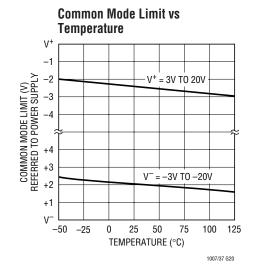


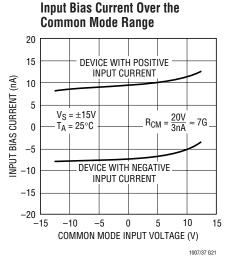


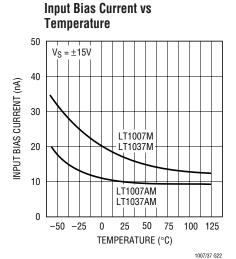


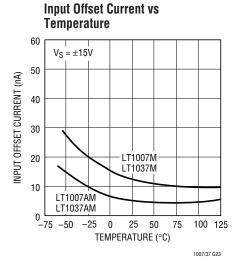
1007/37 G19

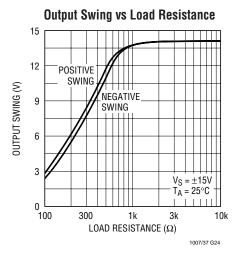
FREQUENCY (Hz)



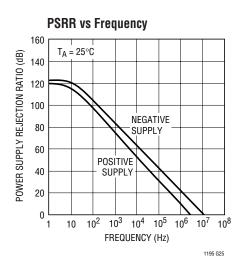


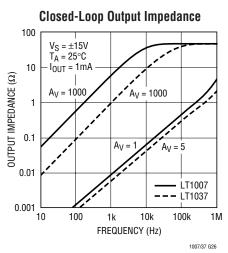


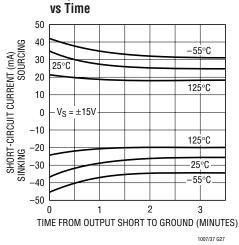




Output Short-Circuit Current

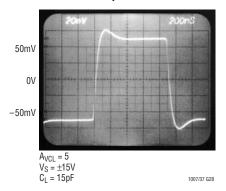




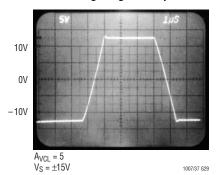




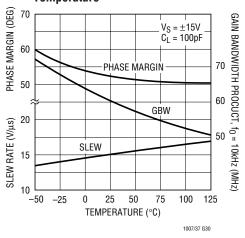
LT1037 Small-Signal Transient Response



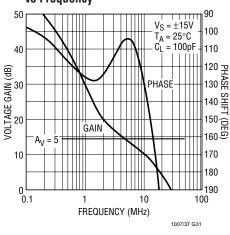
LT1037 Large-Signal Response



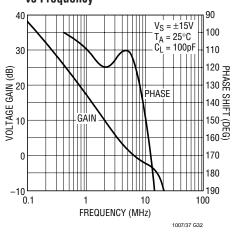
LT1037 Phase Margin, Gain Bandwidth Product, Slew Rate vs Temperature



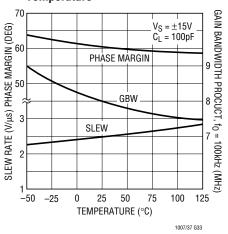
LT1037 Gain, Phase Shift vs Frequency



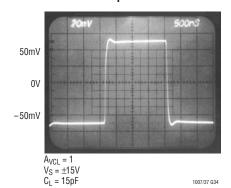
LT1007 Gain, Phase Shift vs Frequency



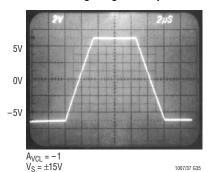
LT1007 Phase Margin, Gain Bandwidth Product, Slew Rate vs Temperature



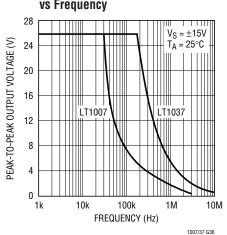
LT1007 Small-Signal Transient Response



LT1007 Large-Signal Response



Maximum Undistorted Output vs Frequency





APPLICATIONS INFORMATION

General

The LT1007/LT1037 series devices may be inserted directly into OP-07, OP-27, OP-37 and 5534 sockets with or without removal of external compensation or nulling components. In addition, the LT1007/LT1037 may be fitted to 741 sockets with the removal or modification of external nulling components.

Offset Voltage Adjustment

The input offset voltage of the LT1007/LT1037 and its drift with temperature, are permanently trimmed at wafer testing to a low level. However, if further adjustment of V_{OS} is necessary, the use of a $10k\Omega$ nulling potentiometer will not degrade drift with temperature. Trimming to a value other than zero creates a drift of $(V_{OS}/300)\mu V/^{\circ}C$, e.g., if V_{OS} is adjusted to $300\mu V$, the change in drift will be $1\mu V/^{\circ}C$ (Figure 1).

The adjustment range with a $10k\Omega$ pot is approximately ± 2.5 mV. If less adjustment range is needed, the sensitivity and resolution of the nulling can be improved by using a smaller pot in conjunction with fixed resistors. The example has an approximate null range of $\pm 200\mu$ V (Figure 2).

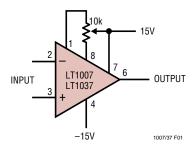


Figure 1. Standard Adjustment

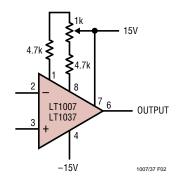


Figure 2. Improved Sensitivity Adjustment

Offset Voltage and Drift

Thermocouple effects, caused by temperature gradients across dissimilar metals at the contacts to the input terminals, can exceed the inherent drift of the amplifier unless proper care is exercised. Air currents should be minimized, package leads should be short, the two input leads should be close together and maintained at the same temperature.

The circuit shown to measure offset voltage is also used as the burn-in configuration for the LT1007/LT1037, with the supply voltages increased to $\pm 20V$ (Figure 3).

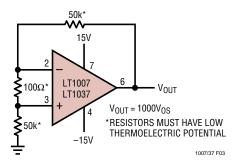


Figure 3. Test Circuit for Offset Voltage and Offset Voltage Drift with Temperature

Unity-Gain Buffer Application (LT1007 Only)

When $R_F \le 100\Omega$ and the input is driven with a fast, large-signal pulse (>1V), the output waveform will look as shown in the pulsed operation diagram (Figure 4).

During the fast feedthrough-like portion of the output, the input protection diodes effectively short the output to the input and a current, limited only by the output short-circuit protection, will be drawn by the signal generator. With $R_F \geq 500\Omega,$ the output is capable of handling the current requirements ($I_L \leq 20 \text{mA}$ at 10V) and the amplifier stays in its active mode and a smooth transition will occur.

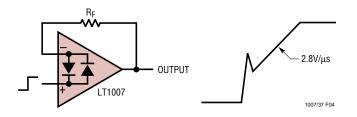


Figure 4. Pulsed Operation



APPLICATIONS INFORMATION

As with all operational amplifiers when $R_F > 2k$, a pole will be created with R_F and the amplifier's input capacitance, creating additional phase shift and reducing the phase margin. A small capacitor (20pF to 50pF) in parallel with R_F will eliminate this problem.

Noise Testing

The 0.1Hz to 10Hz peak-to-peak noise of the LT1007/LT1037 is measured in the test circuit shown (Figure 5a). The frequency response of this noise tester (Figure 5b) indicates that the 0.1Hz corner is defined by only one zero. The test time to measure 0.1Hz to 10Hz noise should not exceed ten seconds, as this time limit acts as an additional zero to eliminate noise contributions from the frequency band below 0.1Hz.

Measuring the typical 60nV peak-to-peak noise performance of the LT1007/LT1037 requires special test precautions:

- 1. The device should be warmed up for at least five minutes. As the op amp warms up, its offset voltage changes typically 3μV due to its chip temperature increasing 10°C to 20°C from the moment the power supplies are turned on. In the ten-second measurement interval these temperature-induced effects can easily exceed tens of nanovolts.
- 2. For similar reasons, the device must be well shielded from air currents to eliminate the possibility of thermo-

- electric effects in excess of a few nanovolts, which would invalidate the measurements.
- 3. Sudden motion in the vicinity of the device can also "feedthrough" to increase the observed noise.

A noise voltage density test is recommended when measuring noise on a large number of units. A 10Hz noise voltage density measurement will correlate well with a 0.1Hz to 10Hz peak-to-peak noise reading since both results are determined by the white noise and the location of the 1/f corner frequency.

Current noise is measured in the circuit shown in Figure 6 and calculated by the following formula:

$$i_n = \frac{\left[\left(e_{no}\right)^2 - \left(130 \text{nV} \bullet 101\right)^2\right]^{1/2}}{\left(1M\Omega\right)\!\left(101\right)}$$

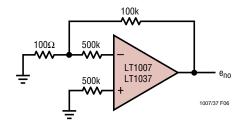


Figure 6

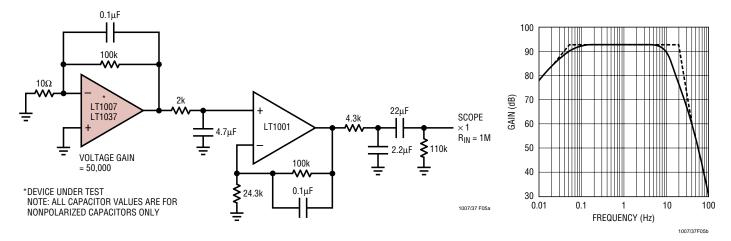


Figure 5a. 0.1Hz to 10Hz Noise Test Circuit

Figure 5b. 0.1Hz to 10Hz Peak-to-Peak Noise Tester Frequency Response



APPLICATIONS INFORMATION

The LT1007/LT1037 achieve their low noise, in part, by operating the input stage at $120\mu\text{A}$ versus the typical $10\mu\text{A}$ of most other op amps. Voltage noise is inversely proportional while current noise is directly proportional to the square root of the input stage current. Therefore, the LT1007/LT1037's current noise will be relatively high. At low frequencies, the low 1/f current noise corner frequency (\approx 120Hz) minimizes current noise to some extent.

In most practical applications, however, current noise will not limit system performance. This is illustrated in the Total Noise vs Source Resistance plot in the Typical Performance Characteristics section, where:

Total Noise = $[(voltage noise)^2 + (current noise • R_S)^2 + (resistor noise)^2]^{1/2}$

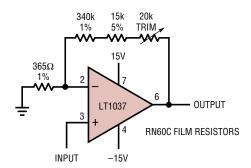
Three regions can be identified as a function of source resistance:

- (i) $R_S \le 400\Omega$. Voltage noise dominates
- (ii) $400\Omega \le R_S \le 50 k$ at 1 kHz Resistor noise $400\Omega \le R_S \le 8 k$ at 10 Hz
- (iii) $R_S > 50k$ at 1kHz Current noise $R_S > 8k$ at 10Hz

Clearly the LT1007/LT1037 should not be used in region (iii), where total system noise is at least six times higher than the voltage noise of the op amp, i.e., the low voltage noise specification is completely wasted.

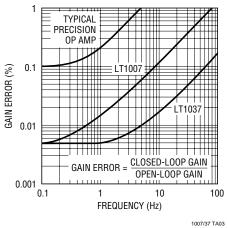
TYPICAL APPLICATIONS

Gain 1000 Amplifier with 0.01% Accuracy, DC to 5Hz



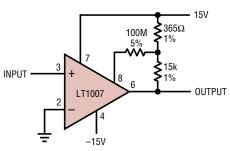
THE HIGH GAIN AND WIDE BANDWIDTH OF THE LT1037 (AND LT1007) IS USEFUL IN LOW FREQUENCY, HIGH CLOSED-LOOP GAIN AMPLIFIER APPLICATIONS. A TYPICAL PRECISION OP AMP MAY HAVE AN OPEN-LOOP GAIN OF ONE MILLION WITH 500kHz BANDWIDTH. AS THE GAIN ERROR PLOT SHOWS, THIS DEVICE IS CAPABLE OF 0.1% AMPLIFYING ACCURACY UP TO 0.3Hz ONLY. EVEN INSTRUMENTATION RANGE SIGNALS CAN VARY AT A FASTER RATE. THE LT1037'S "GAIN PRECISION-BANDWIDTH PRODUCT" IS 200 TIMES HIGHER AS SHOWN.

Gain Error vs Frequency Closed-Loop Gain = 1000



TYPICAL APPLICATIONS

Microvolt Comparator with Hysteresis

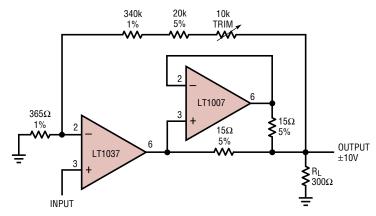


POSITIVE FEEDBACK TO ONE OF THE NULLING TERMINALS CREATES APPROXIMATELY $5\mu V$ OF HYSTERESIS. OUTPUT CAN SINK 16mA.

INPUT OFFSET VOLTAGE IS TYPICALLY CHANGED LESS THAN $5\mu V$ due to the feedback.

1007/37 TA04

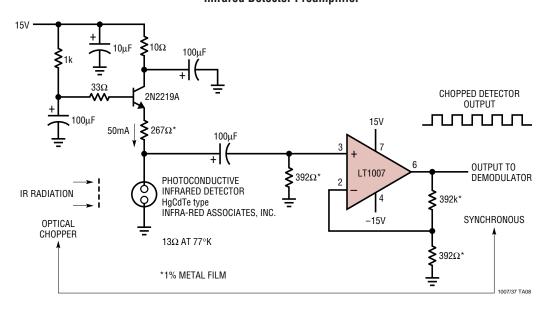
Precision Amplifier Drives 300 Ω Load to $\pm 10V$



THE ADDITION OF THE LT1007 DOUBLES THE AMPLIFIER'S OUTPUT DRIVE TO $\pm 33 \text{ma}$. Gain accuracy is 0.02%, SLIGHTLY DEGRADED COMPARED TO ABOVE BECAUSE OF SELF-HEATING OF THE LT1037 UNDER LOAD.

1007/37 TA05

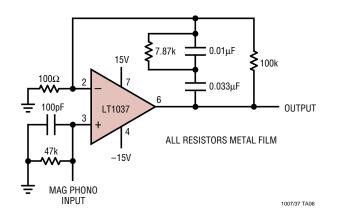
Infrared Detector Preamplifier



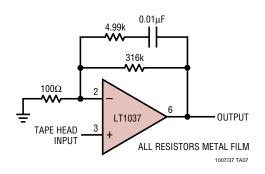
LINEAR TECHNOLOGY

TYPICAL APPLICATIONS

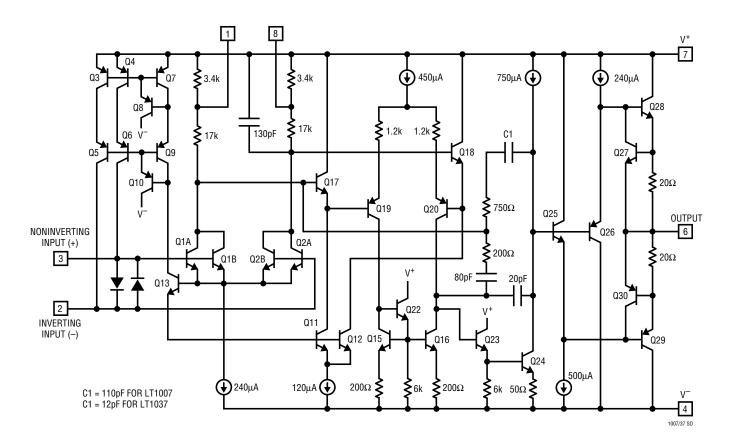
Phono Preamplifier



Tape Head Amplifier

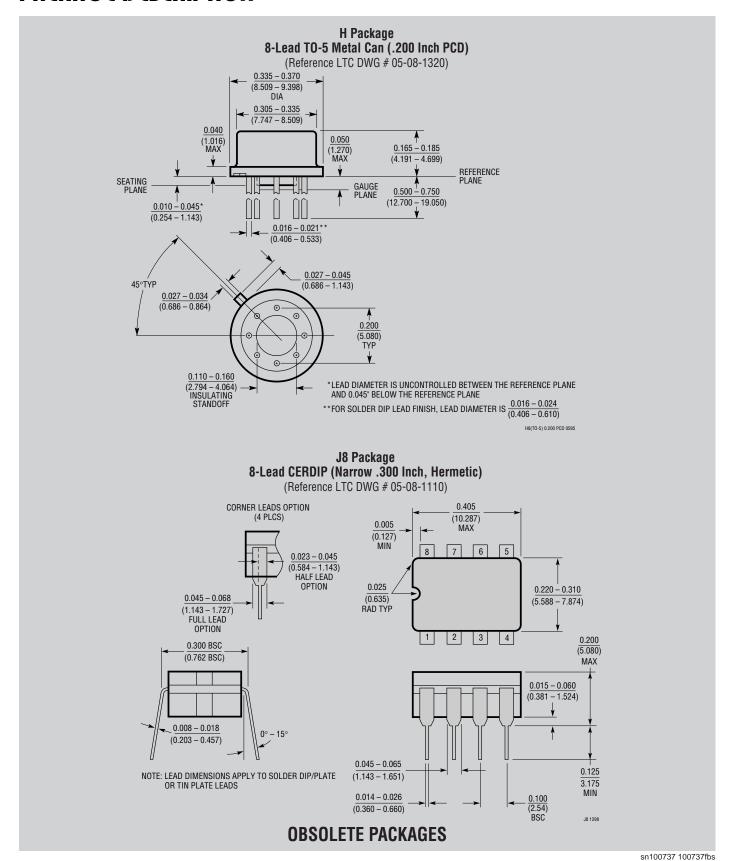


SIMPLIFIED SCHEMATIC





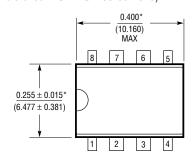
PACKAGE DESCRIPTION

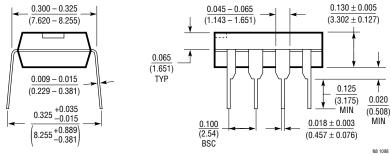


PACKAGE DESCRIPTION

N8 Package 8-Lead PDIP (Narrow .300 Inch)

(Reference LTC DWG # 05-08-1510)

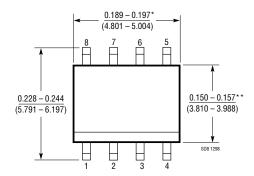


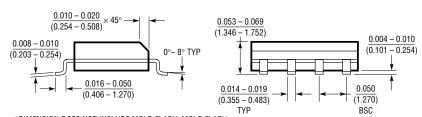


^{*}THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS. MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.010 INCH (0.254mm)

S8 Package 8-Lead Plastic Small Outline (Narrow .150 Inch)

(Reference LTC DWG # 05-08-1610)





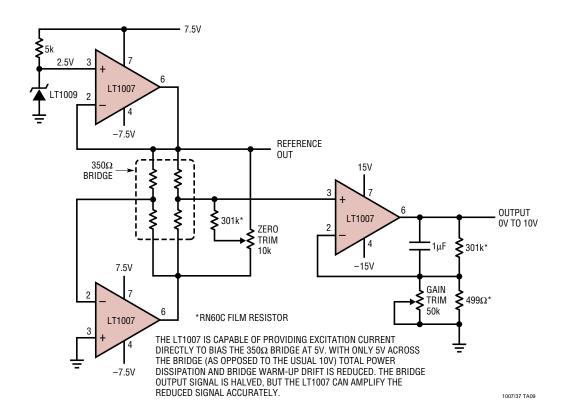
^{*}DIMENSION DOES NOT INCLUDE MOLD FLASH. MOLD FLASH SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

^{**}DIMENSION DOES NOT INCLUDE INTERLEAD FLASH. INTERLEAD FLASH SHALL NOT EXCEED 0.010" (0.254mm) PER SIDE



TYPICAL APPLICATION

Strain Gauge Signal Conditioner with Bridge Excitation



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT1028	Ultralow Noise Precision Op Amp	Lowest Noise 0.85nV/√Hz
LT1115	Ultralow Noise, Low distortion Audio Op Amp	0.002% THD, Max Noise 1.2mV/√Hz
LT1124/LT1125	Dual/Quad Low Noise, High Speed Precision Op Amps	Similar to LT1007
LT1126/LT1127	Dual/Quad Decompensated Low Noise, High Speed Precision Op Amps	Similar to LT1037
LT1498/LT1499	10MHz, 5V/μs, Dual/Quad Rail-to-Rail Input and Output Precision C-Load™ Op Amps	

C-Load is a trademark of Linear Technology Corporation.



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LTC6253CTS8#TRMPBF LT1399HVCS#PBF LT1993CUD-2#PBF LT1722CS8#PBF LT1208CN8#PBF LT1222CN8#PBF

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OP37EN8#PBF LTC6253IMS8#PBF LT1360CS8 OPA2132PAG4 OPA2353UA/2K5 OPA2691I-14D OPA4353UA/2K5 OPA690IDRG4

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