LT1 122

## feATURES

- 100\% Tested Settling Time to 1 mV at Sum Node, 10 V Step
Tested with Fixed Feedback Capacitor
- Slew Rate
- Gain-Bandwidth Product
- Power Bandwidth (20VP-p)
- Unity-Gain Stable; Phase Margin
- Input Offset Voltage
- Input Bias Current
- Input Offset Current
- Low Distortion


## APPLICATIONS

- Fast 12-Bit D/A Output Amplifiers
- High Speed Buffers
- Fast Sample-and-Hold Amplifiers
- High Speed Integrators
- Voltage to Frequency Converters
- Active Filters
- Log Amplifiers
- Peak Detectors
$25^{\circ} \mathrm{C}$
$70^{\circ} \mathrm{C}$
$25^{\circ} \mathrm{C}$
$70^{\circ} \mathrm{C}$


## DESCRIPTIOn

340ns Typ 540ns Max

60V/us Min 14 MHz 1.2 MHz $60^{\circ}$ $600 \mu \mathrm{~V}$ Max 75pA Max 600pA Max
40pA Max
150pA Max

It slews at $80 \mathrm{~V} /$ us and settles in 340 ns. The LT1122 is internally compensated to be unity-gain stable, yet it has a bandwidth of 14 MHz at a supply current of only 7 mA . Its speed makes the LT1122 an ideal choice for fast settling 12-bit data conversion and acquisition systems.
The LT1122 offset voltage of $120 \mu \mathrm{~V}$, and voltage gain of 500,000 also support the 12-bit accurate applications.
The input bias current of 10pA and offset current of 4 pA combined with its speed allow the LT1122 to be used in such applications as high speed sample and hold amplifiers, peak detectors, and integrators.
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## TYPICAL APPLICATION



[^0]Large-Scale Response


## ABSOLUTE MAXIMUM RATINGS (Note 1)

| Supply Voltage ............................................... $\pm 20 \mathrm{~V}$ | Operating Temperature Range |
| :---: | :---: |
| Differential Input Voltage .................................. 40 V | LT1122AM/BM/CM/DM (OBSOLETE).. $-55^{\circ} \mathrm{C}$ to $125^{\circ} \mathrm{C}$ |
| Input Voltage.................................................. 20 V | LT1122AC/BC/CC/DC/CS/DS ............... $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$ |
| Output Short Circuit Duration ....................... Indefinite | Storage Temperature Range |
| Lead Temperature (Soldering, 10 sec.$)$................ $300^{\circ} \mathrm{C}$ | All Devices ..................................... $-65^{\circ} \mathrm{C}$ to $150^{\circ} \mathrm{C}$ |

## pIn COnfiguration




## ORDER INFORMATION

| LEAD FREE FINISH | TAPE AND REEL | PART MARKING | PACKAGE DESCRIPTION | TEMPERATURE RANGE |
| :---: | :---: | :---: | :---: | :---: |
| LT1122ACN8\#PBF | LT1122ACN8\#TRPBF | LT1122ACN8 | 8-Lead Plastic DIP | $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$ |
| LT1122BCN8\#PBF | LT1122BCN8\#TRPBF | LT1122BCN8 | 8-Lead Plastic DIP | $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$ |
| LT1122CCN8\#PBF | LT1122CCN8\#TRPBF | LT1122CCN8 | 8-Lead Plastic DIP | $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$ |
| LT1122DCN8\#PBF | LT1122DCN8\#TRPBF | LT1122DCN8 | 8-Lead Plastic DIP | $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$ |
| LT1122CS8\#PBF | LT1122CS8\#TRPBF | 1122C | 8-Lead Plastic SO | $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$ |
| LT1122DS8\#PBF | LT1122DS8\#TRPBF | 1122D | 8-Lead Plastic SO | $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$ |
| OBSOLETE PACKAGE |  |  |  |  |
| LT1122AMJ8\#PBF | LT1122AMJ8\#TRPBF | LT1122AMJ8 | 8-Lead Hermetic DIP | $-55^{\circ} \mathrm{C}$ to $125^{\circ} \mathrm{C}$ |
| LT1122BMJ8\#PBF | LT1122BMJ8\#TRPBF | LT1122BMJ8 | 8-Lead Hermetic DIP | $-55^{\circ} \mathrm{C}$ to $125^{\circ} \mathrm{C}$ |
| LT1122CMJ8\#PBF | LT1122CMJ8\#TRPBF | LT1122CMJ8 | 8-Lead Hermetic DIP | $-55^{\circ} \mathrm{C}$ to $125^{\circ} \mathrm{C}$ |
| LT1122DMJ8\#PBF | LT1122DMJ8\#TRPBF | LT1122DMJ8 | 8-Lead Hermetic DIP | $-55^{\circ} \mathrm{C}$ to $125^{\circ} \mathrm{C}$ |
| LT1122ACJ8\#PBF | LT1122ACJ8\#TRPBF | LT1122ACJ8 | 8-Lead Hermetic DIP | $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$ |
| LT1122BCJ8\#PBF | LT1122BCJ8\#TRPBF | LT1122BCJ8 | 8-Lead Hermetic DIP | $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$ |
| LT1122CCJ8\#PBF | LT1122CCJ8\#TRPBF | LT1122CCJ8 | 8-Lead Hermetic DIP | $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$ |
| LT1122DCJ8\#PBF | LT1122DCJ8\#TRPBF | LT1122DCJ8 | 8-Lead Hermetic DIP | $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$ |

Consult LTC Marketing for parts specified with wider operating temperature ranges.
Consult LTC Marketing for information on nonstandard lead based finish parts.
For more information on lead free part markings, go to: http://www.linear.com/leadfree/
For more information on tape and reel specifications, go to: http://www.linear.com/tapeandreel/

ELECTRICAL CHARACTERISTICS The $\bullet$ denotes the specifications which apply vere the full operating temperature range, otherwise specifications are at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C} . \mathrm{V}_{\mathrm{S}}= \pm 15 \mathrm{~V}, \mathrm{~V}_{\mathrm{CM}}=0 \mathrm{~V}$ unless otherwise noted. (Note 2)

| SYMBOL | PARAMETER | CONDITIONS | LT1122AM/BM LT1122AC/BC |  |  | LT1122CM/DM LT1122CC/DC LT1122CS/DS |  |  | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN | TYP | MAX | MIN | TYP | MAX |  |
| $\mathrm{V}_{0 S}$ | Input Offset Voltage |  |  | 120 | 600 |  | 130 | 900 | $\mu \mathrm{V}$ |
| Ios | Input Offset Current |  |  | 4 | 40 |  | 5 | 50 | pA |
| $\mathrm{I}_{\mathrm{B}}$ | Input Bias Current |  |  | 10 | 75 |  | 12 | 100 | pA |
|  | Input Resistance Differential Common Mode | $\begin{aligned} & V_{C M}=-10 \mathrm{~V} \text { to } 8 \mathrm{~V} \\ & \mathrm{~V}_{\mathrm{CM}}=8 \mathrm{~V} \text { to } 11 \mathrm{~V} \end{aligned}$ |  | $\begin{aligned} & 10^{12} \\ & 10^{12} \\ & 10^{11} \\ & \hline \end{aligned}$ |  |  | $\begin{aligned} & 10^{12} \\ & 10^{12} \\ & 10^{11} \end{aligned}$ |  | $\Omega$ $\Omega$ $\Omega$ |
|  | Input Capacitance |  |  | 4 |  |  | 4 |  | pF |
| SR | Slew Rate | $A_{V}=-1$ | 60 | 80 |  | 50 | 75 |  | $\mathrm{V} / \mathrm{\mu s}$ |
|  | Settling Time (Note 2) | 10 V to $0 \mathrm{~V},-10 \mathrm{~V}$ to 0 V $100 \%$ Tested: A- and C-Grades to 1 mV at Sum Node B- and D-Grades to 1 mV at Sum Node All Grades to 0.5 mV at Sum Node |  | $\begin{aligned} & 340 \\ & 350 \\ & 450 \end{aligned}$ | 540 |  | $\begin{aligned} & 350 \\ & 360 \\ & 470 \end{aligned}$ | $590$ | ns ns ns |
| GBW | Gain-Bandwidth Product Power Bandwidth | $\mathrm{V}_{\text {OUT }}=20 \mathrm{~V}_{\text {P-P }}$ |  | $\begin{aligned} & 14 \\ & 1.2 \end{aligned}$ |  |  | $\begin{aligned} & \hline 13 \\ & 1.1 \end{aligned}$ |  | $\begin{aligned} & \mathrm{MHz} \\ & \mathrm{MHz} \end{aligned}$ |
| AVOL | Large-Signal Voltage Gain | $\begin{aligned} & V_{\text {OUT }}= \pm 10 \mathrm{~V}, R_{\mathrm{L}}=2 \mathrm{k} \Omega \\ & \mathrm{~V}_{\text {OUT }}= \pm 10 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=600 \Omega \end{aligned}$ | $\begin{aligned} & 180 \\ & 130 \end{aligned}$ | $\begin{aligned} & 500 \\ & 250 \end{aligned}$ |  | $\begin{aligned} & 150 \\ & 110 \end{aligned}$ | $\begin{aligned} & 450 \\ & 220 \end{aligned}$ |  | $\mathrm{V} / \mathrm{mV}$ <br> $\mathrm{V} / \mathrm{mV}$ |
| CMRR | Common-Mode Rejection Ratio | $\mathrm{V}_{\text {CM }}= \pm 10 \mathrm{~V}$ | 83 | 99 |  | 80 | 98 |  | dB |
|  | Input Voltage Range | (Note 4) | $\pm 10.5$ | $\pm 11$ |  | $\pm 10.5$ | $\pm 11$ |  | V |
| PSRR | Power Supply Rejection Ratio | $\mathrm{V}_{S}= \pm 10 \mathrm{~V}$ to $\pm 18 \mathrm{~V}$ | 86 | 103 |  | 82 | 101 |  | dB |
|  | Input Noise Voltage | 0.1 Hz to 10 Hz |  | 3.0 |  |  | 3.3 |  | $\mu \mathrm{V}_{\text {P-P }}$ |
|  | Input Noise Voltage Density | $\begin{aligned} & \mathrm{f}_{0}=100 \mathrm{~Hz} \\ & \mathrm{f}_{0}=10 \mathrm{kHz} \\ & \hline \end{aligned}$ |  | $\begin{aligned} & 25 \\ & 14 \end{aligned}$ |  |  | $\begin{aligned} & 27 \\ & 15 \\ & \hline \end{aligned}$ |  | $\begin{aligned} & \mathrm{nV} / \sqrt{\mathrm{Hz}} \\ & \mathrm{nV} / \sqrt{\mathrm{Hz}} \end{aligned}$ |
|  | Input Noise Current Density | $\mathrm{f}_{0}=100 \mathrm{~Hz}, \mathrm{f}_{0}=10 \mathrm{kHz}$ |  | 2 |  |  | 2 |  | $\mathrm{fA} / \sqrt{\mathrm{Hz}}$ |
| V OUT | Output Voltage Swing | $\begin{aligned} & \mathrm{R}_{\mathrm{L}}=2 \mathrm{k} \Omega \\ & \mathrm{R}_{\mathrm{L}}=600 \Omega \end{aligned}$ | $\begin{gathered} \pm 12 \\ \pm 11.5 \end{gathered}$ | $\begin{gathered} \pm 12.5 \\ \pm 12 \end{gathered}$ |  | $\begin{gathered} \pm 12 \\ \pm 11.5 \end{gathered}$ | $\begin{gathered} \pm 12.5 \\ \pm 12 \end{gathered}$ |  | V |
| IS | Supply Current |  |  | 7.5 | 10 |  | 7.8 | 11 | mA |
|  | Minimum Supply Voltage | (Note 5) | $\pm 5$ |  |  | $\pm 5$ |  |  | V |
|  | Offset Adjustment Range | RPOT $\geq 10 \mathrm{k}$, Wiper to $\mathrm{V}^{+}$ | $\pm 4$ | $\pm 10$ |  | $\pm 4$ | $\pm 10$ |  | mV |

## LT1 122

ELECTRICAL CHARACTERISTICS The • denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $0^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq 70^{\circ} \mathrm{C}$. $\mathrm{V}_{\mathrm{S}}= \pm 15 \mathrm{~V}, \mathrm{~V}_{\mathrm{CM}}=0 \mathrm{~V}$. (Note 2)

| SYMBOL | PARAMETER | CONDITIONS |  | LT1122AC/BC |  |  | LT1122CC/DC <br> LT1122CS/DS |  |  | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | MIN | TYP | MAX | MIN | TYP | MAX |  |
| V ${ }_{\text {OS }}$ | Input Offset Voltage |  | $\bullet$ |  | 350 | 1400 |  | 400 | 2000 | $\mu \mathrm{V}$ |
|  | Average Temperature Coefficient of Input Offset Voltage |  | $\bullet$ |  | 5 | 18 |  | 6 | 25 | $\mu \mathrm{V} /{ }^{\circ} \mathrm{C}$ |
| Ios | Input Offset Current |  | $\bullet$ |  | 12 | 150 |  | 15 | 200 | pA |
| IB | Input Bias Current |  | $\bullet$ |  | 80 | 600 |  | 90 | 800 | pA |
| AVOL | Large-Signal Voltage Gain | $\mathrm{V}_{\text {OUT }}= \pm 10 \mathrm{~V}, \mathrm{R}_{\mathrm{L}} \geq 2 \mathrm{k} \Omega$ | $\bullet$ | 120 | 380 |  | 100 | 340 |  | $\mathrm{V} / \mathrm{mV}$ |
| CMRR | Common-Mode Rejection Ratio | $\mathrm{V}_{\mathrm{CM}}= \pm 10 \mathrm{~V}$ | $\bullet$ | 82 | 98 |  | 78 | 96 |  | dB |
| PSRR | Power Supply Rejection Ratio | $\mathrm{V}_{S}= \pm 10 \mathrm{~V}$ to $\pm 17 \mathrm{~V}$ | $\bullet$ | 84 | 101 |  | 80 | 99 |  | dB |
|  | Input Voltage Range |  | $\bullet$ | $\pm 10$ | $\pm 10.8$ |  | $\pm 10$ | $\pm 10.8$ |  | V |
| V OUT | Output Voltage Swing | $\mathrm{R}_{\mathrm{L}}=2 \mathrm{k} \Omega$ | $\bullet$ | $\pm 11.5$ | $\pm 12.4$ |  | $\pm 11.5$ | $\pm 12.4$ |  | V |
| SR | Slew Rate | $\mathrm{A}_{\mathrm{V}}=-1$ | $\bullet$ | 50 | 70 |  | 40 | 65 |  | $\mathrm{V} / \mathrm{LS}$ |

The odenotes the specifications which apply over the full operating temperature range, otherwise specifications are at $-55^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq 125^{\circ} \mathrm{C} . \mathrm{V}_{\mathrm{S}}= \pm 15 \mathrm{~V}, \mathrm{~V}_{\mathrm{CM}}=0 \mathrm{~V}$. (Note 2)

| SYMBOL | PARAMETER | CONDITIONS |  | LT1122AM/BM |  |  | LT1122CS/DS |  |  | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | MIN | TYP | MAX | MIN | TYP | MAX |  |
| $\mathrm{V}_{\text {OS }}$ | Input Offset Voltage |  | $\bullet$ |  | 650 | 2400 |  | 800 | 3400 | $\mu \mathrm{V}$ |
|  | Average Temperature Coefficient of Input Offset Voltage |  | $\bullet$ |  | 6 | 18 |  | 7 | 25 | $\mu \mathrm{V} /{ }^{\circ} \mathrm{C}$ |
| Ios | Input Offset Current |  | $\bullet$ |  | 0.5 | 6 |  | 0.6 | 9 | nA |
| $\mathrm{I}_{\mathrm{B}}$ | Input Bias Current |  | $\bullet$ |  | 6 | 25 |  | 7 | 35 | nA |
| AVOL | Large-Signal Voltage Gain | $\mathrm{V}_{\text {OUT }}= \pm 10 \mathrm{~V}, \mathrm{R}_{\mathrm{L}} \geq 2 \mathrm{k} \Omega$ | $\bullet$ | 70 | 230 |  | 60 | 200 |  | $\mathrm{V} / \mathrm{mV}$ |
| CMRR | Common-Mode Rejection Ratio | $\mathrm{V}_{\text {CM }}= \pm 10 \mathrm{~V}$ | $\bullet$ | 80 | 97 |  | 76 | 94 |  | dB |
| PSRR | Power Supply Rejection Ratio | $\mathrm{V}_{S}= \pm 10 \mathrm{~V}$ to $\pm 17 \mathrm{~V}$ | $\bullet$ | 83 | 100 |  | 78 | 98 |  | dB |
|  | Input Voltage Range |  | $\bullet$ | $\pm 10$ | $\pm 10.5$ |  | $\pm 10$ | $\pm 10.5$ |  | V |
| $\mathrm{V}_{\text {OUT }}$ | Output Voltage Swing | $\mathrm{R}_{\mathrm{L}}=2 \mathrm{k}$, | $\bullet$ | $\pm 11.3$ | $\pm 12.1$ |  | $\pm 11.3$ | $\pm 12.1$ |  | V |
| SR | Slew Rate | $A_{V}=-1$ | $\bullet$ | 45 | 60 |  | 35 | 55 |  | $\mathrm{V} / \mathrm{\mu s}$ |

The $\bullet$ denotes the specifications which apply over the full operating temperature range, otherwise specifications
are at $-40^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq 85^{\circ} \mathrm{C}$. $\mathrm{V}_{\mathrm{S}}= \pm 15 \mathrm{~V}, \mathrm{~V}_{\mathrm{CM}}=0 \mathrm{~V}$. (Note 6 )

| SYMBOL | PARAMETER | CONDITIONS |  | LT1122AM/BM |  |  | LT1122CS/DS |  |  | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | MIN | TYP | MAX | MIN | TYP | MAX |  |
| $\mathrm{V}_{\text {OS }}$ | Input Offset Voltage |  | $\bullet$ |  | 450 | 1900 |  | 500 | 2700 | $\mu \mathrm{V}$ |
|  | Average Temperature Coefficient of Input Offset Voltage |  | $\bullet$ |  | 6 | 20 |  | 7 | 28 | $\mu \mathrm{V} /{ }^{\circ} \mathrm{C}$ |
| 10 S | Input Offset Current |  | $\bullet$ |  | 30 | 600 |  | 40 | 900 | pA |
| $\mathrm{I}_{\mathrm{B}}$ | Input Bias Current |  | $\bullet$ |  | 230 | 2000 |  | 260 | 2700 | pA |
| AVOL | Large-Signal Voltage Gain | $\mathrm{V}_{\text {OUT }}= \pm 10 \mathrm{~V}, \mathrm{R}_{\mathrm{L}} \geq 2 \mathrm{k} \Omega$ | $\bullet$ | 95 | 340 |  | 80 | 300 |  | $\mathrm{V} / \mathrm{mV}$ |
| CMRR | Common-Mode Rejection Ratio | $\mathrm{V}_{\text {CM }}= \pm 10 \mathrm{~V}$ | $\bullet$ | 80 | 98 |  | 76 | 96 |  | dB |
| PSRR | Power Supply Rejection Ratio | $\mathrm{V}_{S}= \pm 10 \mathrm{~V}$ to $\pm 17 \mathrm{~V}$ | $\bullet$ | 83 | 100 |  | 78 | 98 |  | dB |
|  | Input Voltage Range |  | $\bullet$ | $\pm 10$ | $\pm 10.6$ |  | $\pm 10$ | $\pm 10.6$ |  | V |
| V OUT | Output Voltage Swing |  | $\bullet$ | $\pm 11.3$ | $\pm 12.2$ |  | $\pm 11.3$ | $\pm 12.2$ |  | V |
| SR | Slew Rate | $\mathrm{A}_{V}=-1$ | $\bullet$ | 45 | 60 |  | 35 | 60 |  | $\mathrm{V} / \mathrm{\mu s}$ |
|  |  |  |  |  |  |  |  |  |  | 1122fb |
| $4$ |  | For more information | T11 |  |  |  |  |  | K | $\mathrm{EAR}^{2}$ |

## ELECTRICAL CHARACTERISTICS

Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.
Note 2: The LT1122 is measured in an automated tester in less than one second after application of power. Depending on the package used, power dissipation, heat sinking, and air flow conditions, the fully warmed up chip temperature can be $10^{\circ} \mathrm{C}$ to $50^{\circ} \mathrm{C}$ higher than the ambient temperature.
Note 3: Settling time is $100 \%$ tested for $\mathrm{A}-$ and C -grades using the settling time test circuit shown. This test is not included in quality assurance sample testing.

Note 4: Input voltage range functionality is assured by testing offset voltage at the input voltage range limits to a maximum of 4 mV (A, B grades), to 5.7 mV (C, D grades).
Note 5: Minimum supply voltage is tested by measuring offset voltage to 7 mV maximum at $\pm 5 \mathrm{~V}$ supplies.
Note 6: The LT1122 is not tested and not quality-assurance-sampled at $-40^{\circ} \mathrm{C}$ and at $85^{\circ} \mathrm{C}$. These specifications are guaranteed by design, correlation and/or inference from $-55^{\circ} \mathrm{C}, 0^{\circ} \mathrm{C}, 25^{\circ} \mathrm{C}, 70^{\circ} \mathrm{C}$ and/or $125^{\circ} \mathrm{C}$ tests.

## Settling Time Test Fixture



## TYPICAL PERFORMANCE CHARACTERISTICS



Settling Time
(Input from OV to -10V)


Settling Time
(Input from 10V to OV)


100ns/DIV

Large-Signal Response


Settling Time
(Input from OV to 10V)


100ns/DIV

## Undistorted Output Swing

 vs Frequency

1122 TPC01


Gain, Phase vs Frequency



1122 TPC04

## TYPICAL PERFORMANCE CHARACTERISTICS




TIME AFTER POWER ON (MINUTES)

## Total Harmonic Distortion + Noise

 vs Frequency Inverting Gain

Input Bias and Offset Currents
Over Temperature


1122 TPC06

## Noise Spectrum



FREQUENCY (Hz)
1122 TPC09
Total Harmonic Distortion + Noise vs Frequency Noninverting Gain


1122 TPC11


1122 TPC07

### 0.1 Hz to 10 Hz Noise



1122 TPC10
Intermodulation Distortion
(CCIF Method) vs Frequency
LT1122 and LF156*

*SEE LT1115 DATA SHEET FOR DEFINITION OF CCIF TESTING

## APPLICATIONS INFORMATION

## Settling Time Measurements

Settling time test circuits shown on some competitive devices' data sheets require:

1. A "flat top" pulse generator. Unfortunately, flattop pulse generators are not commercially available.
2. A variable feedback capacitor around the device under test. This capacitor varies over a four-to-one range. Presumably, as each op amp is measured for settling time, the capacitor is fine tuned to optimize settling time for that particular device.
3. A small inductor load to optimize settling.

The LT1122's settling time is 100\% tested in the test circuit shown. No "flat top" pulse generator is required. The test circuit can be readily constructed, using commercially available ICs. Of course, standard high frequency board construction techniques should be followed. All LT1122s are measured with a constant feedback capacitor. No fine tuning is required.

## Speed Boost/Overcompensation Terminal

Pin 8 of the LT1122 can be used to change the input stage operating current of the device. Shorting Pin 8 to the positive supply (Pin 7) increases slew rate and bandwidth by about $25 \%$, but at the expense of a reduction in phase margin by approximately 18 degrees. Unity-gain capacitive load handling decreases from typically 500pF to 100pF.

Conversely, connecting a 15 k resistor from Pin 8 to ground pulls 1 mA out of Pin 8 (with $\mathrm{V}^{+}=15 \mathrm{~V}$ ). This reduces slew rate and bandwidth by $25 \%$. Phase margin and capacitive load handling improve; the latter typically increasing to 800pF.

## High Speed Operation

As with most high speed amplifiers, care should be taken with supply decoupling, lead dress and component placement.

The power supply connections to the LT1122 must maintain a low impedance to ground over a bandwidth of 20 MHz . This is especially important when driving a significant resistive or capacitive load, since all current delivered to the load comes from the power supplies. Multiple high quality bypass capacitors are recommended for each power supply line in any critical application. A $0.1 \mu \mathrm{~F}$ ceramic and a $1 \mu \mathrm{~F}$ electrolytic capacitor, as shown, placed as close as possible to the amplifier (with short lead lengths to power supply common) will assure adequate high frequency bypassing, in most applications.


When the feedback around the op amp is resistive ( $\mathrm{R}_{\mathrm{F}}$ ), a pole will be created with $R_{F}$, the source resistance and capacitance ( $\mathrm{R}_{\mathrm{S}}, \mathrm{C}_{\mathrm{S}}$ ), and the amplifier input capacitance ( $\mathrm{C}_{\mathrm{IN}} \approx 4 \mathrm{pF}$ ). In low closed-loop gain configurations and with $R_{S}$ and $R_{F}$ in the kilohm range, this pole can create excess phase shift and even oscillation. A small capacitor $\left(C_{F}\right)$ in parallel with $R_{F}$ eliminates this problem. With $R_{S}\left(C_{S}+C_{I N}\right)=R_{F} C_{F}$, the effect of the feedback pole is completely removed.


## TYPICAL APPLICATIONS

Quartz Stabilized Oscillator With 9ppm Distortion


## PACKAGE DESCRIPTION

Please refer to http://www.linear.com/designtools/packaging/ for the most recent package drawings.

## N Package

8-Lead PDIP (Narrow . 300 Inch)
(Reference LTC DWG \# 05-08-1510 Rev I)


NOTE:
NOTE:

1. DIMENSIONS ARE $\frac{\text { INCHES }}{\text { MILLIMETERS }}$
*THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS.
MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED . 010 INCH ( 0.254 mm )

## PACKAGE DESCRIPTION

Please refer to http://www.linear.com/designtools/packaging/ for the most recent package drawings.

## J8 Package

3-Lead CERDIP (Narrow . 300 Inch, Hermetic)
(Reference LTC DWG \# 05-08-1110)


OBSOLETE PACKAGE

## PACKAGE DESCRIPTION

Please refer to http://www.linear.com/designtools/packaging/ for the most recent package drawings.

S8 Package
8-Lead Plastic Small Outline (Narrow . 150 Inch)
(Reference LTC DWG \# 05-08-1610 Rev G)


## REVISIO HISTORY (Revision history begins at Rev B)

| REV | DATE | DESCRIPTION | PAGE NUMBER |
| :---: | :---: | :--- | :---: |
| B | $02 / 14$ | Updated data sheet to current standards. New Order Information Table, Package Descriptions | $2,10-12$ |

## TYPICAL APPLICATION

Wide-Band, Filtered, Full Wave Rectifier



OUTPUT DC = RMS VALUE OF INPUT
BANDWIDTH WITH $10 \mathrm{~V}_{\text {P-P }}$ INPUT $=2 \mathrm{MHz}$

## RELATED PARTS

| PART NUMBER | DESCRIPTION | COMMENTS |
| :--- | :--- | :--- |
| LT1022 | High Speed Precision JFET Op Amp | $23 \mathrm{~V} / \mu \mathrm{s}$ Min Slew Rate, $250 \mu \mathrm{~V} \mathrm{~V}_{\text {OS }}$ |
| LT1055/LT1056 | Precision High Speed JFET Op Amps | $16 \mathrm{~V} / \mu$ s Slew Rate, $150 \mu \mathrm{~V} \mathrm{~V}_{\text {OS }}$ |
| LT1464 | 1 MHz C-LoadTM Stable JFET Op Amp | Capacitive Loads Up to 10 nF |
| LTC $^{\oplus} 6244$ | 50 MHz Low Noise CMOS Op Amp | 1 1pA I $\mathrm{I}_{\mathrm{B}}, 100 \mu \mathrm{~V}$ Max $\mathrm{V}_{0 S}, 1.5 \mu \mathrm{~V}_{\mathrm{p}-\mathrm{p}, 0.1 \mathrm{~Hz}}$ to 10 Hz Noise |

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[^0]:    12-BIT CURRENT OUTPUT D/A CONVERTER
    $\mathrm{C}_{\mathrm{F}}=5 \mathrm{pF}$ T0 17pF
    (DEPENDING ON D/A CONVERTER USED)

