

LTC1049

FEATURES

- Low Supply Current: 200µA
- No External Components Required
- Maximum Offset Voltage: 10µV
- Maximum Offset Voltage Drift: 0.1µV/°C
- Single Supply Operation: 4.75V to 16V
- Input Common Mode Range Includes Ground
- Output Swings to Ground
- Typical Overload Recovery Time: 6ms
- Available in 8-Pin SO and PDIP Packages

APPLICATIONS

- 4mA to 20mA Current Loops
- Thermocouple Amplifiers
- Electronic Scales
- Medical Instrumentation
- Strain Gauge Amplifiers
- High Resolution Data Acquisition

Low Power Zero-Drift Operational Amplifier with Internal Capacitors **DESCRIPTION**

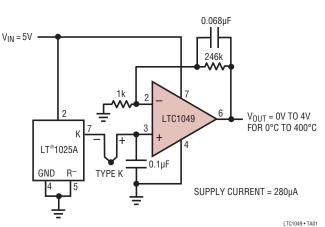
The LTC®1049 is a high performance, low power zero-drift operational amplifier. The two sample-and-hold capacitors usually required externally by other chopper stabilized amplifiers are integrated on the chip. Further, the LTC1049 offers superior DC and AC performance with a nominal supply current of only 200µA.

The LTC1049 has a typical offset voltage of 2μ V, drift of 0.02 μ V/°C, 0.1Hz to 10Hz input noise voltage of 3μ V_{P-P} and typical voltage gain of 160dB. The slew rate is 0.8V/ μ s with a gain bandwidth product of 0.8MHz.

Overload recovery time from a saturation condition is 6ms, a significant improvement over chopper amplifiers using external capacitors.

The LTC1049 is available in a standard 8-pin plastic dual in line, as well as an 8-pin SO package. The LTC1049 can be a plug-in replacement for most standard op amps with improved DC performance and substantial power savings.

TYPICAL APPLICATION

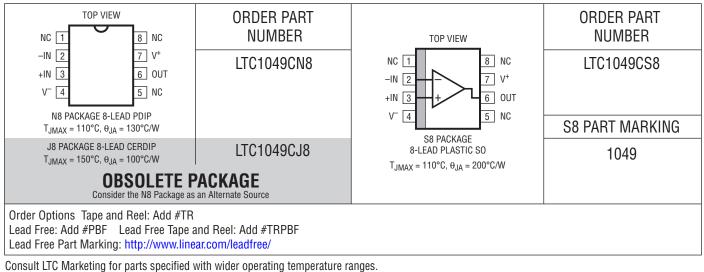


Single Supply Thermocouple Amplifier

LINEAR TECHNOLOGY

ABSOLUTE MAXIMUM RATINGS (Note 1)

PACKAGE/ORDER INFORMATION



ELECTRICAL CHARACTERISTICS The • denotes the specifications which apply over the full operating

temperature range, otherwise specifications are at $T_A = 25^{\circ}C$. $V_S = \pm 5V$, unless noted.

PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
Input Offset Voltage	(Note 3)			±2	±10	μV
Average Input Offset Drift	(Note 3)	•		±0.02	±0.1	μV/°C
Long Term Offset Voltage Drift				50		nV√mo
Input Offset Current		•		±30	±100 ±150	pA pA
Input Bias Current		•		±15	±50 ±150	pA pA
Input Noise Voltage	0.1Hz to 10Hz 0.1Hz to 1Hz			3 1		μV _{P-P} μV _{P-P}
Input Noise Current	f = 10Hz (Note 4)			2		fA√Hz
Common Mode Rejection Ratio	V _{CM} = V ⁻ to 2.7V	•	110	130		dB
Power Supply Rejection Ratio	$V_{S} = \pm 2.375 V \text{ to } \pm 8 V$	•	110	130		dB
Large-Signal Voltage Gain	$R_{L} = 100 k\Omega, V_{OUT} = \pm 4.75 V$	•	130	160		dB
Maximum Output Voltage Swing	$R_L = 10k\Omega$	•	-4.6/3.2	-4.9/4.2		V V
	$R_L = 100 k\Omega$	•	±4.9	±4.97		V
Slew Rate	$R_L = 10k\Omega, C_L = 50pF$			0.8		V/µs
Gain Bandwidth Product				0.8		MHz
Supply Current	No Load	•		200	330 495	μA μA
Internal Sampling Frequency				700		Hz
	·	· · · ·	÷			1049fb



ELECTRICAL CHARACTERISTICS

Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

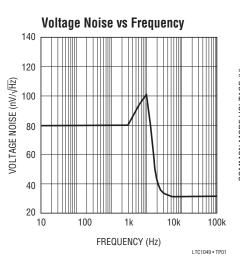
Note 2: Connecting any terminal to voltages greater than V⁺ or less than V⁻ may cause destructive latch-up. It is recommended that no sources operating from external supplies be applied prior to power-up of the LTC1049.

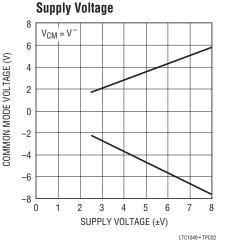
Note 3: These parameters are guaranteed by design. Thermocouple effects preclude measurement of these voltage levels in high speed automatic test systems. V_{OS} is measured to a limit determined by test equipment capability.

Note 4: Current Noise is calculated from the formula:

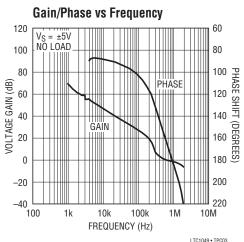
 $I_N = \sqrt{(2q \bullet I_b)}$ where $q = 1.6 \bullet 10^{-19}$ Coulomb.

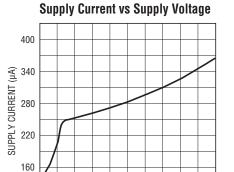
TYPICAL PERFORMANCE CHARACTERISTICS





Common Mode Input Range vs





9

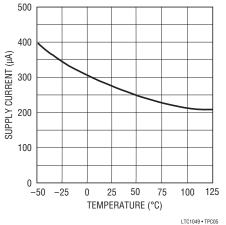
TOTAL SUPPLY VOLTAGE (V)

10 11 12

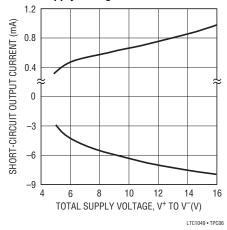
13 14 15

LTC1049 • TPC04

Supply Current vs Temperature



Output Short-Circuit Current vs Supply Voltage



1049fb



100

5 6 7 8

TYPICAL PERFORMANCE CHARACTERISTICS

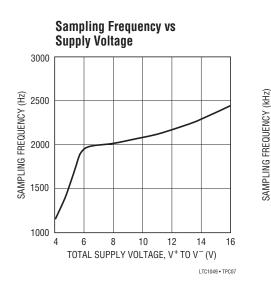
400mV

OV OUTPUT

-5V

LTC1049 • TPC10

INPUT



Overload Recovery

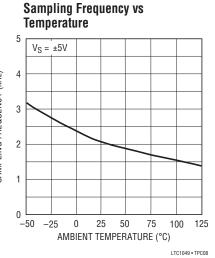
0.2V/DIV

2V/DIV

 $A_{V} = -100$

 $V_{S} = \pm 5V$

0.5ms/DIV



Response

 $A_V = 1$ $R_L = 10k$

 $C_L = 50 pF$ $V_S = \pm 5V$

100mV

1µs/DIV

STEP



LTC1049 • TPC11



100

FREQUENCY (Hz)

1k

10k

100k

1 TC1049 • TPC09

CMRR vs Frequency

 $V_{S} = \pm 5V$

160

140

120 100

80

60

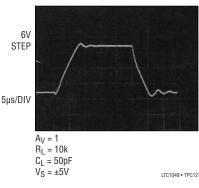
40

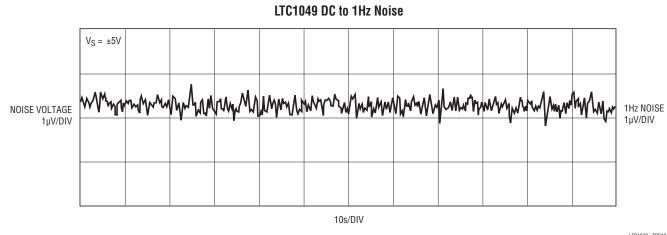
20

0

10

CMRR (dB)



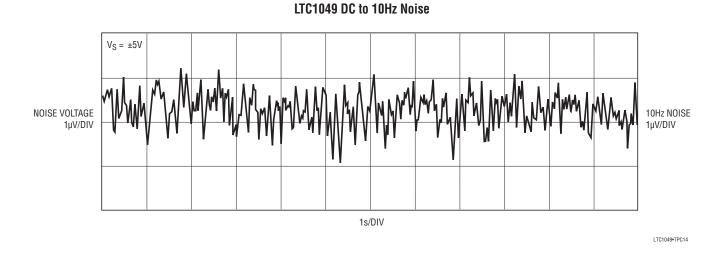


LTC1049 • TPC13

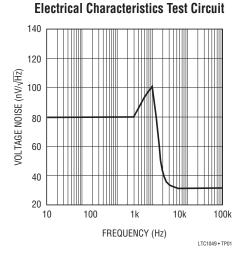


4

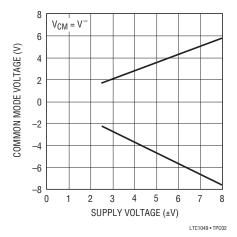
TYPICAL PERFORMANCE CHARACTERISTICS



TEST CIRCUITS



DC to 10Hz and DC to 1Hz Noise Test Circuit



LINEAR TECHNOLOGY

APPLICATIONS INFORMATION

ACHIEVING PICOAMPERE/MICROVOLT PERFORMANCE

Picoamperes

In order to realize the picoampere level of accuracy of the LTC1049, proper care must be exercised. Leakage currents in circuitry external to the amplifier can significantly degrade performance. High quality insulation should be used (e.g., Teflon[™], Kel-F); cleaning of all insulating surfaces to remove fluxes and other residues will probably be necessary—particularly for high temperature performance. Surface coating may be necessary to provide a moisture barrier in high humidity environments.

Board leakage can be minimized by encircling the input connections with a guard ring operated at a potential close to that of the inputs: in inverting configurations, the guard ring should be tied to ground; in noninverting connections, to the inverting input. Guarding both sides of the printed circuit board is required. Bulk leakage reduction depends on the guard ring width.

Microvolts

Thermocouple effects must be considered if the LTC1049's ultralow drift is to be fully utilized. Any connection of dissimilar metals forms a thermoelectric junction producing an electric potential which varies with temperature (Seebeck effect). As temperature sensors, thermocouples exploit this phenomenon to produce useful information. In low drift amplifier circuits the effect is a primary source of error.

Connectors, switches, relay contacts, sockets, resistors, solder, and even copper wire are all candidates for thermal EMF generation. Junctions of copper wire from different manufacturers can generate thermal EMFs of $200nV/^{\circ}C$ —twice the maximum drift specification of the LTC1049. The copper/kovar junction, formed when wire or printed circuit traces contact a package lead, has a thermal EMF of approximately $35\mu V/^{\circ}C$ —300 times the maximum drift specification of the LTC1049.

Minimizing thermal EMF-induced errors is possible if judicious attention is given to circuit board layout and component selection. It is good practice to minimize the number of junctions in the amplifier's input signal path. Avoid connectors, sockets, switches, and relays where possible. In instances where this is not possible, attempt to balance the number and type of junctions so that differential cancellation occurs. Doing this may involve deliberately introducing junctions to offset unavoidable junctions.

PACKAGE-INDUCED OFFSET VOLTAGE

Package-induced thermal EMF effects are another important source of errors. It arises at the copper/kovar junctions formed when wire or printed circuit traces contact a package lead. Like all the previously mentioned thermal EMF effects, it is outside the LTC1049's offset nulling loop and cannot be cancelled. The input offset voltage specification of the LTC1049 is actually set by the package-induced warm-up drift rather than by the circuit itself. The thermal time constant ranges from 0.5 to 3 minutes, depending on package type.

LOW SUPPLY OPERATION

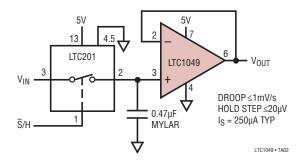
The minimum supply for proper operation of the LTC1049 is typically below 4.0V (\pm 2.0V). In single supply applications, PSRR is guaranteed down to 4.7V (\pm 2.35V) to ensure proper operation down to the minimum TTL specified voltage of 4.75V.

PIN COMPATIBILITY

The LTC1049 is pin compatible with the 8-pin versions of 7650, 7652 and other chopper-stabilized amplifiers. The 7650 and 7652 require the use of two external capacitors connected to Pins 1 and 8 which are not needed for the LTC1049. Pins 1, 5, and 8 of the LTC1049 are not connected internally; thus, the LTC1049 can be a direct plug- in for the 7650 and 7652, even if the two capacitors are left on the circuit board.



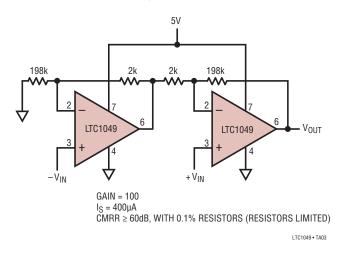
TYPICAL APPLICATIONS



Low Power, Low Hold Step Sample-and-Hold



TYPICAL APPLICATIONS

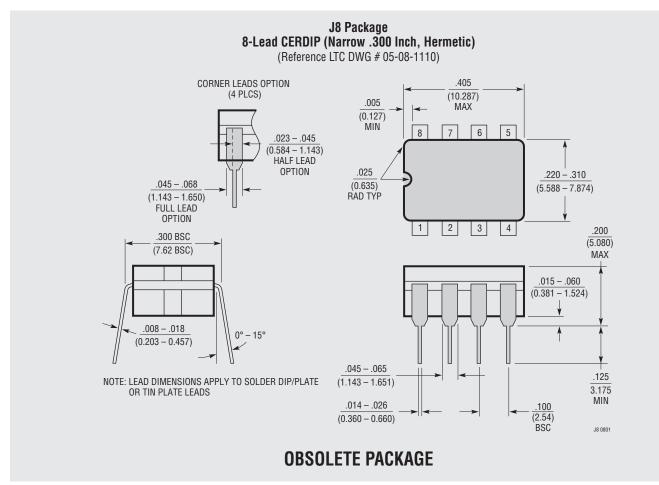


Low Power, Single Supply, Low Offset Instrumentation Amp



1049fb

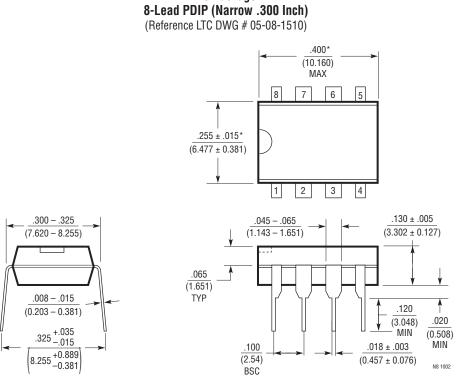
PACKAGE DESCRIPTION





1049fb

PACKAGE DESCRIPTION



N8 Package

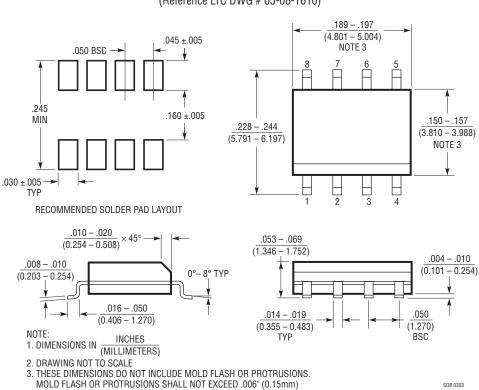
NOTE: INCHES 1. DIMENSIONS ARE MILLIMETERS

*THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS. MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED .010 INCH (0.254mm)



1049fb

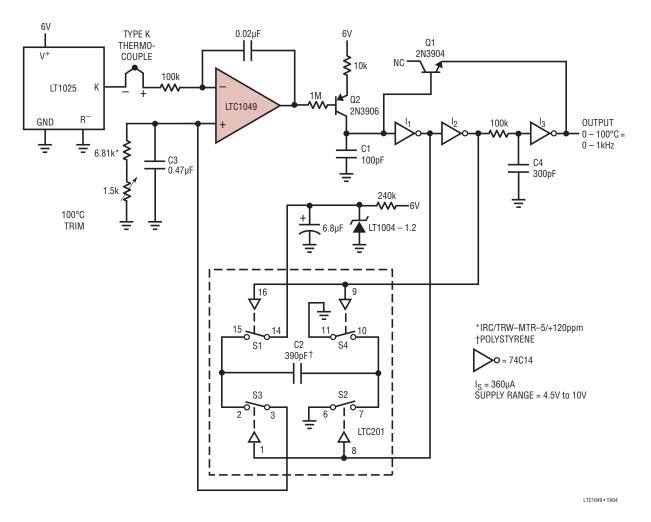
PACKAGE DESCRIPTION



S8 Package 8-Lead Plastic Small Outline (Narrow .150 Inch) (Reference LTC DWG # 05-08-1610)



TYPICAL APPLICATION



Thermocouple-Based Temperature to Frequency Converter



X-ON Electronics

Largest Supplier of Electrical and Electronic Components

Click to view similar products for Precision Amplifiers category:

Click to view products by Analog Devices manufacturer:

Other Similar products are found below :

561681F LT6005HGN#PBF LT6238CGN#PBF LT6238HGN#PBF OP05CN8#PBF OP227GN#PBF LT6020IDD-1#PBF LT1124CS8#TR NCV20166SN2T1G NCS21802MUTBG LT1637MPS8 LT1498IS8 LT1492CS8 TLC27L7CP TLV2473CDR LMP2234AMA/NOPB LMP7707MA/NOPB 5962-8859301M2A LMP2231AMAE/NOPB LMP2234AMTE/NOPB LMC6022IM/NOPB LMC6024IM/NOPB LMC6081IMX/NOPB LMP2011MA/NOPB LMP2231AMFE/NOPB LMP2232BMA/NOPB LMP2234AMAE/NOPB LMP7717MAE/NOPB LMV2011MA/NOPB TLC2201AMDG4 TLE2024BMDWG4 TLV2474AQDRG4Q1 TLV2472QDRQ1 TLC4502IDR TLC27M2ACP TLC2652Q-8DG4 OPA2107APG4 TL054AIDR TLC272CD AD8539ARMZ LTC6084HDD#PBF LT1638CMS8#TRPBF LTC1050CN8#PBF LT1112ACN8#PBF LT1996AIDD#PBF LT1112CN8#PBF LTC6087CDD#PBF LT1078S8#PBF LT1079ACN#PBF LTC6242HVCDHC#PBF