## ProxSense ${ }^{\circledR}$ IQS263B Datasheet

## 3 Channel Capacitive Touch and Proximity Controller with 8-bit Resolution Slider or Scroll Wheel

The IQS263B ProxSense ${ }^{\circledR}$ IC is a 3-channel projected (or self) capacitive proximity and touch controller with best in class sensitivity, signal to noise ratio and power consumption. Other features include automatic tuning for sense electrodes, internal reference capacitor and internal regulator to reduce total system cost.

## Main Features

- 3 Self or Mutual Channel Capacitive Controller
- Configurable 8 -bit $2 / 3$ channel slider or 3 channel scroll wheel
- Up to 80 Hz report rate
- On chip Movement Detection algorithm
- SAR compliance in mobile devices according to the IEC 62209-2 ed1.0 standard and the FCC standard (KDB 616217 - D04 SAR for laptop and tablets v01)
- Left and right flick gesture recognition
- Automatic adjustment for optimal performance (ATI)
- User selectable Proximity and Touch thresholds
- Long proximity range
- Automatic drift compensation
- Fast I ${ }^{2} \mathrm{C}$ Interface
- Event mode or Streaming modes
- Low Power, suitable for battery applications

- Supply voltage: 1.8 V to 3.6 V
- $\quad<3 \mu \mathrm{~A}$ Active sensing in LP mode
- $<100 \mu \mathrm{~A}$ full running current up to 40 Hz report rate


## Applications

- SAR detection for tablets and mobile phones
- SAR detection for wireless charging stations (operating frequency $>100 \mathrm{MHz}$ )
- SAR detection for USB dongle-type devices
- Sliders/Scroll wheels for remote controls
- Movement detection devices (anti-theft)
- Volume Control \& Lid closure detection for laptops
- MP3 players
- Portable Electronics

Available Options

- Wearable Electronics

| TA | MSOP10 | DFN10 (3x3) |
| :---: | :---: | :---: |
| $-20^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}$ | IQS263B | IQS263B |

## Contents

1 INTRODUCTION ..... 3
2 ANALOGUE FUNCTIONALITY ..... 3
3 DIGITAL FUNCTIONALITY ..... 3
4 PASSING THE SAR QUALIFICATION TESTING ..... 3
5 HARDWARE CONFIGURATION ..... 4
6 USER CONFIGURABLE OPTIONS ..... 10
7 SLIDERS/WHEEL COORDINATE TO POSITION MAPPING ..... 18
8 PROXSENSE MODULE ..... 22
9 COMMUNICATION ..... 27
10 COMMUNICATION COMMAND/ADDRESS STRUCTURE ..... 29
11 SPECIFICATIONS ..... 42
12 PACKAGING INFORMATION ..... 45
13 DEVICE MARKING ..... 51
14 ORDERING INFORMATION ..... 52

## Revision History

| Version | Description | Date |
| :--- | :--- | :--- |
| 1.00 | First Release | August 2017 |
| 1.01 | Second Release | November 2017 |
| 1.02 | Third Release | December 2017 |
| 1.03 | Fourth Release | March 2018 |
| 1.04 | Fifth Release | October 2018 |
| 1.05 | Sixth Release | August 2019 |
| 1.06 | ESD rating changed from 8 kV to 4kV, page 41. |  |

## Functional Overview

## 1 Introduction

The IQS263B is a 3 channel projected (or self) capacitive proximity and touch sensor featuring an internal voltage regulator and reference capacitor (Cs).

The device has 4 possible pins for the connection of the sense electrodes, which consist of 3 Self electrodes, or 3 receivers and 1 transmitter. Three pins are used for serial data communication through the $I^{2} \mathrm{C}^{\top \mathrm{M}}$ compatible protocol, including an optional RDY pin. The transmitter pin can also be used as an I/O for output upon proximity detection when the device is used in Self capacitive mode.

The device automatically tracks slow varying environmental changes via various filters, detects noise and is equipped with an Automatic Tuning Implementation (ATI) to adjust the device for optimal sensitivity.

### 1.1 Applicability

All specifications, except where specifically mentioned otherwise, provided by this datasheet are applicable to the following ranges:

- Temperature $-20^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$
- Supply voltage (VDDHI) 1.8 V to 3.6 V


### 1.2 Device Operation

With a normal activation (hand brought close) the output bit(s) will become active. These outputs will clear as soon as the action is reversed (hand taken away). In addition a separate movement output will become active when movement is detected according to a movement threshold. Movement may be detected before the normal touch threshold is crossed. Movement detection is done via a completely separate digital filter.

## 2 Analogue Functionality

CRX and CTX electrodes are arranged in a suitable configuration that results in a mutual capacitance ( Cm ) between the two electrodes.

CTX is charged up to a set positive potential during a charge cycle which results in a negative charge buildup at CRX.

The resulting charge displacement is then measured within the IQS263B device through a charge transfer process that is periodically initiated by the digital circuitry. The capacitance measurement circuitry makes use of an internal reference capacitor $\mathrm{C}_{\mathrm{s}}$ and voltage reference ( $V_{\text {REF }}$ ).

The measuring process is referred to as a conversion and consists of the discharging of Cs and Cx capacitors, the charging of Cx and then a series of charge transfers from Cx to Cs until a trip voltage is reached. The number of charge transfers required to reach the trip voltage is referred to as the Counts (CS) value.

The analogue circuitry further provides functionality for:

- Power On Reset (POR) detection.
- Brown Out Detection (BOD).
- Internal regulation provides for accurate sampling.


## 3 Digital Functionality

The digital processing functionality is responsible for:

- Managing BOD and WDT events.
- Initiation of conversions at the selected rate.
- Processing of CS and execution of algorithms.
- Monitoring and execution of the ATI algorithm.
- Signal processing and digital filtering.
- Detection of PROX and TOUCH events.
- Managing outputs of the device.
- Managing serial communications.

4 Passing the SAR

IQ Switch ${ }^{\circledR}$
ProxSense ${ }^{\circledR}$ Series

## Qualification Testing

The IQS263B offers a movement monitoring algorithm, to help the host controller distinguish between long term user activations and stuck conditions.

For design information on passing the SAR qualification testing with Azoteq movement based sensors, please see the application note: "AZD073 SAR qualification with Azoteq movement-based sensors.

## 5 Hardware Configuration

The IQS263B can be configured to charge in Self- or Projected-Capacitive mode through the

### 5.1 IQS263B Pin Out



Figure 5.1 IQS263B Pin out in MSOP-10.

| GND | 1 |  | 10 | CTX/PO |
| :---: | :---: | :---: | :---: | :---: |
| CRX1 | 2 |  | 9 | CRX3 |
| CRX2 | 3 | 263 | 8 | SCL |
| VDDHI | 4 |  | 7 | SDA |
| VREG | 5 |  | 6 | RDY |

Figure 5.2 IQS263B Pin out in DFN10.
Table 5.1 IQS263B MSOP-10 \& DFN10 Pin-outs.

| Pin | Self-Capacitive | Projected-Capacitive | Function |
| :---: | :---: | :---: | :---: |
| 1 | GND | GND | Ground |
| 2 | CX1 | $\mathrm{R} \times 1$ | Sensor Electrode |
| 3 | CX2 | Rx2 | Sensor Electrode |
| 4 | VDDHI | VDDHI | Power Supply |
| 5 | VREG | VREG | Regulated Output |
| 6 | RDY | RDY | Communication Interrupt |
| 7 | SDA | SDA | Serial Data |
| 8 | SCL | SCL | Serial Clock |
| 9 | CX3 | Rx3 | Sensor Electrode |
| 10 | PO | Tx | Prox Out/Sensor Electrode |

5.2 Reference Design

### 5.2.1 Projected Schematic



Figure 5.3 IQS263B projected capacitive reference design.

### 5.2.2 Projected Electrodes



Figure 5.4 Example layout of projected wheel OR slider that wraps around - 0 to 191 coordinates available.

### 5.2.3 Example Routing



Figure 5.5 Example routing of a projected slider. GND on top layer shields the Rx lines from the user on the bottom layer.


Figure 5.6 Example routing of 3 discrete buttons. These buttons are suggested for application requiring high noise immunity.

### 5.2.4 Projected Setup

For both the wheel and slider shown in Figure 5.4 the IQS263B slider setup in ProxSettings1 must be set to the "Wheel" selection (Bit3:2 as 10).

The slider setup of the IQS263B in ProxSettings1 for the 3 discrete keys such as shown in Figure 5.6 can be set to "3CH Slider" selection (Bit3:2 as 11). Since the buttons have no interleaving, the coordinate tracking will not be linear. However, recognising flick events will still be achieved.

### 5.2.5 Self Schematic



Figure 5.7 IQS263B self-capacitive reference design.

### 5.2.6 Self Electrodes



Figure 5.8 Example of a 3 Channel Self capacitive wheel and slider. Suggested gaps between electrodes $\geq 0.8 \mathrm{~mm}$.

IQ Switch ${ }^{\circledR}$


Figure 5.9 Examples of 2 Channel Self capacitive sliders. Suggested gaps between electrodes $\geq 0.8 \mathrm{~mm}$.

### 5.2.7 Self Setup

For both the wheel and slider shown in Figure 5.8 the IQS263B slider setup in ProxSettings1 must be set to the "Wheel" selection (Bit3:2 as 10).
For the slider shown in Figure 5.9 the IQS263B slider setup in ProxSettings1 must be set to the "2CH slider" selection (Bit3:2 as 01).

### 5.3 Power Supply and PCB Layout

Azoteq IC's provide a high level of on-chip hardware and software noise filtering and ESD protection (refer to application note "AZD013 - ESD Overview'). Designing PCB's with better noise immunity against EMI, FTB and ESD in mind, it is always advisable to keep the critical noise suppression components like the de-coupling capacitors and series resistors in Figure 5.3 as close as possible to the IC. Always maintain a good ground connection and ground pour underneath the IC. For more guidelines please refer to the relevant application notes as mentioned in Section 5.4.
Where a system level ESD strike is found to cause the IC to go into ESD induced latch-up, it is suggested that the supply current to the IQS263B IC is limited by means of a series resistor that could limit the maximum supply current to the IC to $<80 \mathrm{~mA}$.

### 5.4 Design Rules for Harsh EMC Environments


> Applicable application notes: AZD013, AZD015, AZD051, and AZD052.

### 5.5 High Sensitivity

Through patented design and advanced signal processing, the device is able to provide extremely high sensitivity to detect proximity. This enables designs to detect proximity at distances that cannot be equaled by most other products. When the device is used in environments where high levels of noise or floating metal objects exist, a reduced proximity threshold is proposed to ensure reliable functioning of the sensor. The high sensitivity also allows the device to sense through overlay materials with low dielectric constants, such as wood or porous plastics.
For more guidelines on the layout of capacitive sense electrodes, please refer to application note AZD008, available on the Azoteq web page: www.azoteq.com.

## 6 User Configurable Options

The IQS263B requires configuration by a master/host controller or MCU. The user needs to select the number of channels and corresponding touch and proximity thresholds.

### 6.1 Setup

### 6.1.1 Self and Projected Modes

The IQS263B can function as a Self or Projected capacitive controller, depending on the application requirement. The IQS263B is default in Self capacitive mode. Pre-configured devices can be ordered for projected mode, or the user can select the operation mode as projected during initialization by setting the "Proj Mode" bit in the System Flags register (0x01H; byte 0).

### 6.1.2 Global Filter Halt

By default the IQS263B implements a global filter halt. This realizes that all LTA filters are frozen upon any proximity or touch detection. This optimizes sensitivity by reducing the risk of desensitizing of touch channels during user operation, especially for slider functions.

It is, however, possible to disable the global filter halt, whereby each channel will halt its own LTA upon a detection of an undebounced event on that channel. To disable the global filter halt, the user must set the "Ind Halt" bit in the System Flags register $(0 \times 01 \mathrm{H}$; byte 0$)$.

### 6.2 ProxSetting0

### 6.2.1 Disable ATI

The IQS263B has the ability to automatically retune sensor electrodes when the counts drift outside a pre-defined ATI band. This allows the IQS263B to keep optimal sensitivity during different environment. To disable the feature, the "ATI OFF" bit needs to be set in the ProxSettings0 register ( $0 \times 09$; byte 0). Disabling this feature only disables
the automatic retuning; the MCU can at any time still force retuning with the Redo-ATI command.

### 6.2.2 Partial ATI

If it is required to have the ATI time reduced, the IQS263B can use partial ATI by setting the "ATI Partial" bit in the ProxSettings0 register (0x09; byte 0). The designer must also specify the sensitivity multiplier (option 1 to 4 ) as the IQS263B will only calculate the compensation multiplier and compensation. The Partial ATI option reduces start-up and re-tuning times, but does require the designer to verify that the base values achieved are within the desired range.

### 6.2.3 ATI Band

The user has the option to select the retuning band as $1 / 4$ of the ATI target (default is $1 / 8$ of the ATI target) if it is desired to have a wider range for the counts to drift with environmental change before the device retune the electrodes. The wider band is achieved by setting the "ATI BAND" bit in the ProxSettings0 register (0x09; byte 0).

### 6.2.4 Stream ATI

In order to facilitate faster start-up and retuning times, the communication windows are stopped during ATI on the IQS263. If the designer would like to be able to read data after every charge cycle during ATI, the communication can be enabled by setting the "Stream ATI" bit in the ProxSettings0 register ( $0 \times 09$; byte 0). A communication window can still be forced by the MCU with a RDY handshake (pulling the RDY line low) at any time even if the "Stream ATI" bit is not set.

### 6.2.5 Oscillator

The IQS263B response times can be increased (at the cost of higher current consumption) by setting the internal oscillator at 4 MHz (default 2 MHz ). The faster speed is enabled by setting the " 4 MHz " bit in the ProxSettings0 register (0x09; byte 0).

### 6.2.6 Redo-ATI

The IQS263B can be forced to ATI at any time, regardless of present events. To force retuning set the "Redo ATI" bit in the ProxSettings0 register (0x09; byte 0). The "Redo ATI" bit will automatically clear after having been set.

### 6.2.7 Reseed

The IQS263B LTA filters can be reseeded to the count values at any time to clear any output event. If count values are outside the ATI band, retuning will be triggered. To reseed set the "Reseed" bit in the ProxSettings0 register (0x09; byte 0). The "Reseed" bit will automatically clear after having been set.
Setting the Reseed bit will shift all LTA filters to a value of LTA ${ }_{\text {new }}=C S+8$ (CS -8 for Self). The LTA will then track the CS value until they are even.

Performing a reseed action on the LTA filters, will effectively clear any proximity and/or touch conditions that may have been established prior to the reseed call.

### 6.2.8 Force Halt

The user has the option to halt the LTA to avoid any reseed or re-tuning events from taking place. This can be used in situations where the counts are expected to go in the wrong direction as a result of a controlled action in the application. To freeze the LTA filters set the "Force Halt" bit in the ProxSettings0 register (0x09; byte 0).

### 6.3 ProxSettings1

### 6.3.1 WDT Disable

The WDT (watchdog timer) is used to reset the IC if a problem (for example a voltage spike) occurs during communication. The WDT will time-out (and thus reset the device) after twot if no valid communication occurred during this time.

The WDT can be disabled during development by setting the "WDT Off" bit in the ProxSettings1 register (0x09; byte 1). It
is not recommended to disable the WDT for production.

### 6.3.2 Event Mode

By default the device operates in full streaming mode. There is an option for an event-driven ${ }^{2} \mathrm{C}$ communication mode (also called "Event Mode"), with the RDY pin indicating a communication window ONLY after a prescribed event has occurred.

These events include:

- Flicks (left and/or right)
- Tap
- Movement
- ATI
- Slider
- Touch
- Proximity

The RDY pin will indicate events in the following manner:

1. Flick/Swipe Detected: Single RDY low on flick left or right
2. Tap Detected: Single RDY Iow
3. Movement occurring: Single RDY low on movement event
4. ATI: RDY low on ATI start \& again on ATI completing
5. Slider or Wheel: RDY pin low after completion of every charge cycle while a touch remains detected on a wheel/slider channel.
6. Touch: RDY low on each touch entry and exit occurring.
7. Prox: RDY low on entry and exit

If the wheels/sliders are enabled, the device will stream data continuously (after every charge cycle) when a touch is present on one of the wheel/slider channels, even if Event Mode is enabled.

Event Mode can be enabled by setting the "Event Mode" bit in the ProxSettings1 register (0x09; byte 1).

Note: The device is also capable of functioning without a RDY line on a polling basis.

### 6.3.3 LTA Beta

The speed at which the LTA will follow the counts when no event is present (no filter halt) can be changed by adjusting the beta values for the LTA filter. Four options are available by setting the "LTA Beta" bits in the ProxSettings1 register (0x09; byte 1).

### 6.3.4 Slider

The IQS263B can be used to do 8-bit coordinate calculations to implement a slider or a scroll wheel. The wheel requires all 3 touch channels, where the slider can be selected as a 2 (channel $1 \& 2$ required) or 3 channel slider. By default, no calculations are done. To enable a wheel or slider, set the "Slider" bits in the ProxSettings1 register (0x09; byte 1). When using the wheel setting, Deltas are not available in register $0 \times 06$.

### 6.3.5 Count Filter

The count filter is implemented to provide better stability of Counts (CS) in electrically noisy environments.

The filter also enforces a longer minimum sample time for detecting proximity events on CHO , which will result in a slower response rate when the device enters low power modes. The filter can be disabled.
The count filter is implemented on all channels, to aid in the slider coordinate calculations, but touch events are (by default) determined on unfiltered count values.

The count filter can be disabled, or the speed (amount of filtering) adjusted by setting the "CF" bits in the ProxSettings1 register (0x09; byte 1).

### 6.4 ProxSettings2

### 6.4.1 Output Pin

The IQS263B when operating in Self Capacitive mode has an output/input (PO) pin available (pin 10). As output (active low),
it can be set to trigger on proximity, touch (on channel 1) or movement events. As input, it can be used as a synchronization pin to be used as a ZC (zero cross input) for AC applications. To switch between input and output set the "OUT" bits in the ProxSettings2 register (0x09; byte 2). When used as an input pin, the IQS263B will monitor pin 10, and wait for a change in state (either high to low or low to high), before sensing is initiated. This will happen before each cycle of scanning all the keys.

### 6.4.2 Movement

The IQS263B can implement a movement filter on CH 0 or CH 3 . Movement detection will clear/restart the filter halt timer. Movement will be sensed in both directions. By monitoring movement (of the count values), the host controller can distinguish between user activation and stuck conditions.
The host could choose to implement movement on CH 3 or CH 0 , by setting the "Movement" bits in the ProxSettings2 register (0x09; byte 2).

### 6.4.3 Wake 2 Directions

The IQS263B has the ability to wake from low power in both directions of count movement (of the proximity threshold). This could be used to sense release events from low power mode. To enable sensing in both directions, set the "Wake
Release" bit in the ProxSettings2 register (0x09; byte 2).

### 6.4.4 Wheel/Slider Filter

If the IQS263B implements a wheel or slider, the coordinate filter can be disabled for faster output (if the layout and environment allows stable output). To disable the coordinate filter set the "Wheel Filter" bit in the ProxSettings2 register (0x09; byte 2).

### 6.4.5 Force Sleep

MCU can command a sleep during any situation. This sleep can be cancelled by the MCU, or will automatically clear when the LP channels counts differs more than the
proximity threshold from the LTA (in any direction if $\underline{\text { Wake } 2 \text { Directions is enabled). To }}$ force the IQS263B into low power mode set the "Force Sleep" bit in the ProxSettings2 register (0x09; byte 2).

### 6.4.6 Sleep Halt

The IQS263B can go back into low power mode rather than reseed or re-tune (ATI) when a stuck condition or prolonged event is present. A low power time greater than zero need to be specified for this setting. To set up the sleep on halt time out feature set the "Sleep Halt" " bit in the ProxSettings2 register (0x09; byte 2).

### 6.5 ProxSettings3

### 6.5.1 Touch Debounce

The IQS263B can set the touch output on 4 samples where the delta exceeds the touch threshold (default 2 samples). A lower debounce will speed up the response rate, especially for applications using the filtered counts for touch output. To set the touch debounce to 4 set the "Touch Debounce" bit in the ProxSettings3 register (0x09; byte 3).

### 6.5.2 Off Mode

The IQS263B has the option to switch the device off during inactive states of operation. The device will wake up again with activity on the SDA line. Enabling "Off Mode" switches the device off internally, as a result, no register values are maintained. Upon exiting "Off Mode" the standard startup routine is followed, the "Show Reset" bit is set and the device must be re-initialized. This is suitable for applications that require no device operation during defined operation states and is useful when the master has no control over the VDDHI line of the IQS263B. To switch to off mode set the "Off Mode" bit in the ProxSettings3 register (0x09; byte 3).

### 6.5.3 Projected Bias

The IQS263B has the option to change the bias current of the transmitter during projected sensing mode. A larger bias current is required when using larger electrodes, but will also increase the IC

### 6.5.7 Filtered Touches

The IQS263B offers applications that are sensitive to touch detection and releasing upon soft touch events the option to trigger touch outputs on filtered count values. This improves stability, but could reduce response rate. To enable the filtered touches feature set the "Filtered Touches" bit in the ProxSettings3 register (0x09; byte 3).

### 6.5.8 Charge Transfer Frequency

The IQS263B can reduce the charge transfer frequency for applications that require extra sensitivity (for example very thick overlays). The charge transfer frequency can be halved (for both Self and Projected operation) by setting the "Xfer" bit in the ProxSettings3 register (0x09; byte 3).

### 6.6 Active Channels

The IQS263B can be configured to have up to 3 active touch channels ( $\mathrm{CH} 1-\mathrm{CH} 3$ ) with one additional proximity channel ( CH 0 ). By default CHO is a distributed proximity channel, comprised of charging all the channels together in one timeslot.

### 6.7 Proximity Threshold

A proximity threshold for channel 0 can be selected for the application, to obtain the desired proximity trigger level. The proximity threshold is selectable between 1 (most sensitive) and 255 (least sensitive) counts. These threshold values (i.e. 1-255) are specified in Counts (CS) in the Thresholds register ( $0 \times 0 \mathrm{~A}$; byte 0 ). The default proximity threshold is four counts.

### 6.8 Touch Thresholds

A touch threshold for each channel can be selected by the designer to obtain the desired touch sensitivity and is selectable between $1 / 256$ (most sensitive) to $255 / 256$ (least sensitive). The touch threshold is calculated as a fraction of the Long-Term

$$
T_{T H R}=x / 256 \times L T A
$$

With lower target values (therefore lower LTA's) the touch threshold will be lower and vice versa.

Individual touch thresholds can be set for each channel, by writing to the Thresholds register ( $0 \times 0$; ; byte 1) for channel 1, byte 2 for channel 2 and byte 3 for channel 3 . The default touch threshold is $16 / 256$ times the LTA.

### 6.9 Adjustable Report Rate

The IQS263B can adjust its report rate from to 80 Hz ( 4 Mhz oscillator) to $40 \mathrm{~Hz}(2 \mathrm{MHz})$. The IQS263B can also disable turbo mode, which will yield a fixed sampling period, to optimize the count filter operation.

### 6.10 Gestures

The IQS263B implements a tap gesture on the slider or scroll wheel. Left and right flick gestures are only implemented on the slider (no double tap). For flick events, one of the slider options should be active.

### 6.10.1 Tap

The IQS263B implements an adjustable TAP timer for configurable tap recognition. The timer is started once a touch event is recognised. A tap event will be registered if the touch event is released (cleared) before the timer reaches the adjustable tap time. The tap time can be adjusted in the Gesture Timer register ( $0 \times 0 \mathrm{C}$, byte 0).

### 6.10.2 Flicks

The IQS263B implements an adjustable Flick recognition on the slider setups. The user has the option to specify the flick threshold in the Gesture Timer register ( $0 \times 0 \mathrm{C}$, byte 1) to set the amount that the coordinates must be adjusted by the user before the flick event is set. There is also the option to adjust the time in which the coordinate must move by setting the flick timer in the Gesture Timer register ( $0 \times 0 \mathrm{C}$, byte 2).

### 6.11 Halt times

The LTA filter for all channels will halt on proximity or touch events. The halt time is specified through the memory map in terms of number of samples (sampling of all active channels) by specifying a value in the Thresholds register (0x0A; byte 6). The halt time is calculated as this value times 50. Writing FF will result in "Always Halt" while 00 will be seen as never halt. The default value is 20 , therefore the halt time will be

$$
t_{H A L T}=20 \times 50
$$

which is 1000 samples. The halt timer clears on any new touch, proximity or slider activity.

### 6.12 Power Modes

### 6.12.1 LP Modes

The IQS263B IC has a configurable low power mode, specifically designed to reduce current consumption for low power and battery applications. When designing for low power operation, the $\mathrm{V}_{\text {REG }}$ capacitors should ensure that $\mathrm{V}_{\text {reg }}$ does not drop more than 50 mV during low power operations.

The power modes are implemented around the occurrence of a charge cycle every $t_{L P}$ seconds. The value of $t_{L P}$ is determined by the custom ( $L P_{\text {value }}$ ) value between 1 and

255 , multiplied by 16 ms . The $\mathrm{LP}_{\text {value }}$ can be specified by writing to the Timings register (0x0B; byte 0). It is suggested not to exceed $L P_{\text {value }}=32$ ( 512 ms ) as the reduction in current consumption is negligible, response time becomes notable slower, and special attention needs to be paid to $V_{\text {REG }}$ capacitor design to ensure stable operation. Only CHO is charged during LP, and is forced active ( CH 0 cannot be disabled). The other active channels will be periodically charged to keep their LTA filter values up to date.

Lower sampling times yield significantly lower power consumption (but also decreases the response time of the first event).

NOTE: While in any power mode the device will zoom to Boost Power (BP) mode whenever the condition (CS - LTA) ${ }^{1}$ > PROX_TH or TOUCH_TH holds, indicating a possible proximity or touch event. This improves the response time. The device will remain in BP mode for 4 seconds and then return to the selected power mode. The Zoom function allows reliable detection of events with counts being produced at the BP rate. The LP charge cycle timing is illustrated in Figure 6.1.


[^0]IQ Switch ${ }^{\circledR}$
ProxSense ${ }^{\circledR}$ Series
Figure 6.1 IQS263B Charge Cycle Timing in Low Power Mode.
Table 6.1 Typical Timings

| Typical timings of IQS263B in Low Power |  |  |
| :--- | :--- | :--- |
| $\mathrm{t}_{\text {sense }}$ | 1.1 | ms |
| $\mathrm{t}_{\text {process }}$ | 1.9 | ms |
| $\mathrm{t}_{\text {comms }}$ | Application specific | ms |
| Scan Period | Should not exceed 512ms |  |

### 6.13 Base Value

The IQS263B has the option to select the base value of each channel individually during the Full ATI algorithm. This provides the user with another option to select the sensitivity of the IQS263B without any changes required in the hardware (CRX/CTX sizes and routing, etc).

The base values are set by writing to the Multipliers register ( $0 \times 07$; bytes 4 and 5 ). There are 16 different options to choose from. To choose a custom base value, select partial ATI.

The base value influences the overall sensitivity of the channel and establishes a base count from where the ATI algorithm starts executing. A lower base value will typically result in a higher sensitivity of the respective channel, as lower multipliers will be selected, and more compensation would be required.

### 6.14 Target Value

The default target value of the IQS263B is 384 counts for the touch channels and 512 counts for the proximity channel.

The target values are calculated by multiplying the value in the Timings \&

Targets register (0x0B; byte 1) for channel 1 to channel 3 and byte 2 for channel 0 by 8.

Example: CH 0 target $=$ Register Value $\times 8$ $=64$ (default) $\times 8=512$.

### 6.15 Additional Features

### 6.15.1 Force Halt

The Force Halt bit can be set to halt all current LTA values and prevent them from being adjusted towards the CS values.

Setting this bit overrides all filter halt settings and prevents the device from performing reATI events in cases where the CS values persist outside the ATI boundaries for extended periods of time. Reseed will also not be possible.

### 6.15.2 Halt Timeout Reseed Block

After the halt timer times out upon an event, the IQS263B will reseed all channels and possibly redo ATI if any counts are outside the allowed band. The use has the option to block this reseed on any of the touch channels (not CH 0 ) allowing that channel(s) to keep its output state and effectively have an always halt condition. Note that other channels could drift outside the ATI band, but no ATI will take place if a channel is halted in this way. To set any combination of channels up for this state, select the corresponding bits in the Halt Timeout Reseed Block register (0x0A, byte 5).

## 7 Sliders/Wheel Coordinate to Position Mapping

The coordinate mapping from user finger positions given in this section is applicable to both Self- and Projected Capacitive sensing, even though examples of both will not be shown.
The absolute coordinate is only calculated when the IQS263B is in "Wheel" mode. This is done by setting Bits (3:2) in ProxSettings1 (0x09; Byte 1) to 0b10. For "2CH Slider" mode (0b01) and "3CH Slider" mode (0b11) only the relative coordinate is calculated. Sliders can, however, be used in "Wheel" mode to enable an absolute coordinate calculation, see section 7.4 for more details.

### 7.1 Two (2) Channel Slider

The 2-channel slider is recommended for applications requiring only the relative coordinate and/or the flick event. The 2 channel slider is not optimal for absolute coordinate values, therefore, the absolute coordinate is not calculated in this mode and reading the Wheel 1 Low byte at the Wheel Coordinates (0x02; Byte 0) address will return a value of 0 . Reading the Relative Coordinate Low and Relative Coordinate High bytes will return the relative coordinate value as normal.

If an absolute coordinate is required from a 2 -channel slider, it can easily be calculated by reading the Deltas $(0 \times 06)$ for channel 1 and 2 and applying the following formula:

$$
\begin{equation*}
\text { Coordinate }=\frac{\left(0 \cdot \Delta_{1}\right)+\left(\max \cdot \Delta_{2}\right)}{\left(\Delta_{1}+\Delta_{2}\right)} \tag{eq. 7.1}
\end{equation*}
$$

Where "max" is a user defined value designating the highest possible coordinate value on the slider. Figure 7.1 shows the channel numbers and finger positions for the main definable positions where channels overlap on a 2 channel slider.


Figure 7.1 Finger positions for coordinate mapping shown on a 2 channel slider.
Table 7.1 shows the approximate coordinate that equation 7.1 will produce for the finger positions shown in Figure 7.1.

Table 7.1 2CH slider coordinate mapping.

| Positions | Coordinates |
| :---: | :---: |
| A | 0 |
| A to B | 1 to $[(\max / 2)-1]$ |
| B | $\max / 2$ |
| B to C | $[(\max / 2)+1]$ to $[\max -1]$ |
| C | $\max$ |

7.2 Three (3) Channel Slider

The 3 channel slider is ideal for applications requiring flick events and a relative coordinate. Once again, the absolute coordinate is not calculated for the 3 channel slider. However, the MCU can easily calculate the absolute coordinate by reading the Deltas ( $0 \times 06$ ) and using the following formula:

$$
\begin{equation*}
\text { Coordinate }=\frac{\left(0 \cdot \Delta_{1}\right)+\left((\max / 2) \cdot \Delta_{2}\right)+\left(\max \cdot \Delta_{3}\right)}{\left(\Delta_{1}+\Delta_{2}+\Delta_{3}\right)} \tag{eq. 7.2}
\end{equation*}
$$

Where "max" is a user defined value designating the highest possible coordinate value on the slider. Figure 7.2 shows the channel numbers and finger positions for the main definable positions where channels overlap on a 3 channel slider.


Figure 7.2 Finger positions for coordinate mapping shown on a 3-channel slider.

Table 7.2 shows the approximate coordinate that equation 7.2 will produce for the finger positions shown in Figure 7.3.

Table 7.2 3CH slider coordinate mapping.

| Positions | Coordinates |
| :---: | :---: |
| A | 0 |
| A to B | 1 to $[(\max / 4)-1]$ |
| B | $\max / 4$ |
| B to C | $[(\max / 4)+1]$ to $[(\max / 2)-1]$ |
| C | $\max / 2$ |
| C to D | $[(\max / 2)+1] \operatorname{to}[(3 / 4 \max )-1]$ |
| D | $3 / 4 \max$ |
| D to E | $[(3 / 4 \max )+1] \operatorname{to}[\max -1]$ |
| E | $\max$ |

### 7.3 Wheel / Wrap around Slider

In "Wheel" mode the IQS263B will calculate the coordinate position automatically, the MCU can then simply read the value from Wheel 1 Low byte, Wheel Coordinates (0x02; Byte 0). Figure 7.3 shows the channel numbers and finger positions for the main definable positions where channels overlap on a wraparound slider and wheel.


Figure 7.3 Finger positions for coordinate mapping shown on a wheel and wrap around slider.

Table 7.3 shows the coordinate that the IQS263B will report for the finger positions shown in Figure 7.3.

Table 7.3 Wrap around slider or wheel coordinate mapping.

| Positions | Coordinates |
| :---: | :---: |
| A | 0 |
| A to B | 1 to 31 |
| B | 32 |
| B to C | 33 to 63 |
| C | 64 |
| C to D | 65 to 96 |
| D | 96 |
| D to E | 97 to 127 |
| E | 128 |
| E to F | 129 to 159 |
| F | 160 |
| F to A | 161 to 191 |

As seen from figure 7.3 and table 7.3, the coordinate value increases linearly from a minimum value of 0 to a maximum value of 191 . For the wheel, the coordinate will wrap around from 191 back to 0 and vice versa when scrolling through position $A$.

### 7.4 Non-Wrap Around Sliders in Wheel Mode

Both 2 channel and 3 channel non-wrap around sliders can also be implemented in wheel mode. This will allow a coordinate value to be read from the IQS263B instead of having to calculate it with the MCU. However, this will reduce the coordinate range as shown below.

For a 2 channel slider the coordinate will range from 0 to 64 as shown in figure 7.4 and table 7.4 below.


| Positions | Coordinates |
| :---: | :---: |
| A | $0 / 191$ |
| A to B | 1 to 31 |
| B | 32 |
| B to C | 33 to 63 |
| C | 64 |

Table 7.4 Coordinate mapping for a 2 channel slider in Wheel mode.

Figure 7.4 Finger positions for coordinate mapping shown on a 2 channel slider in Wheel mode.

For a 3 channel slider the coordinate will range from 0 to 127 as shown in figure 7.5 and table 7.5 below.


Figure 7.5 Finger positions for a 3 channel slider in Wheel mode.

| Positions | Coordinates |
| :---: | :---: |
| A | $0 / 191$ |
| A to B | 1 to 31 |
| B | 32 |
| B to C | 33 to 63 |
| C | 64 |
| C to D | 65 to 95 |
| D | 96 |
| D to E | 97 to 126 |
| E | 127 |

Table 7.5 Coordinate mapping for a 3 channel slider in Wheel mode.

It should be noted that non-wrap around 2 channel and 3 channel sliders in wheel mode will sometimes output a value of 191 instead of 0 when touched at the zero end of the slider. Provision must be made in the MCU code for this.

## 8 ProxSense ${ }^{\circledR}$ Module

The IQS263B contains a ProxSense ${ }^{\circledR}$ module that uses patented technology to provide detection of proximity and touch conditions on numerous sensing lines.

The ProxSense ${ }^{\circledR}$ module is a combination of hardware and software, based on the principles of charge transfer measurements.

### 8.1 Charge Transfer Concept

On ProxSense ${ }^{\circledR}$ devices like the IQS263, capacitance measurements are taken with a charge transfer process that is periodically initiated.

For projected capacitive sensing, the device measures the capacitance between 2 electrodes referred to as the transmitter (CTX) and receiver (CRX).

The measuring process is referred to as a charge transfer cycle and consists of the following:

- Discharging of an internal sampling capacitor ( $\mathrm{C}_{\mathrm{s}}$ ) and the electrode capacitors (mutual: CTX \& CRX) on a channel.
- charging of CTX's connected to the channel
- and then a series of charge transfers from the CRX's to the internal sampling capacitors ( $\mathrm{C}_{\mathrm{s}}$ ), until the trip voltage is reached.

The number of charge transfers required to reach the trip voltage on a channel is referred to as the Current Samples (CS) or Count value (Counts).

The device continuously repeats charge transfers on the sense electrodes connected to the CRX pins. For each channel a Long Term Average (LTA) is calculated (12 bit unsigned integer values). The count (CS) values (12 bit unsigned integer values) are
processed and compared to the LTA to detect Touch and Proximity events.

Please note: Attaching scope probes to the CTX/CRX pins will influence the capacitance of the sense electrodes and therefore the related CS values of those channels. This will have an instant effect on the CS measurements.

### 8.2 Rate of Charge Cycles

The IQS263B samples all its active channels (up to $3+$ channel 0 for proximity) in 4 timeslots. The charge sequence (as measured on the receive electrodes) is shown in Figure 8.1, where CHO , the Proximity channel, charges first, followed by all other active channels. There is only a communication window after all active channels have been charged.

The charging of CHO comprises the simultaneous charging of the three receive electrodes (CRX0, CRX1 and CRX2) in conjunction with the transmit electrode, thus realising a distributed load mutual capacitive sense electrode.

In self-capacitive mode, CHO is also a distributed channel charging all 3 CX channels together.

### 8.2.1 Boost Power rate

With the IQS263B zoomed to Boost Power (BP) mode, the active channels are charged at a fixed sampling period ( $\mathrm{t}_{\text {sAMPLE }}$ ) per channel (if Turbo Mode is not enabled). This is done to ensure regular samples for processing of results, and fix timings for the halt times.

It is calculated as each channel having a time tsample $=$ charge/conversion time ( $t_{\text {SENSE }}$ ) + computation time ( $\mathrm{t}_{\text {PRocess }}$ ) of approximately $\mathrm{t}_{\text {sAMPLE }}=1.6 \mathrm{~ms}$. Thus the time between consecutive samples on a specific channel (Scan Period) will depend on the number of enabled channels and the length of communication between the IQS263B and the host MCU.

Communication will always happen after processing of channel 0 . Due to processing and charging happening in parallel, the first active channel (default channel 1) will charge while channel 0 is processed. Therefore, communication windows will always be after the first active channel has sense process
completed conversions. The IQS263B does check for MCU requesting a communication window after ever channel completed charging.


Figure 8.1 IQS263B Charge Sequence timing diagram in Boost Power mode.

Table 8.1 Typical Timings

| Typical timings of IQS263 |  |  |
| :--- | :--- | :--- |
| $t_{\text {sense }}$ | 840 | $\mu \mathrm{~s}$ |
| $t_{\text {process }}$ | 3.9 | ms |
| $t_{\text {comms }}$ | 2 | ms |
| Scan Period ${ }^{1}$ | 22 | ms |

Typical timings of the charge sequence shown above are listed in Table 8.1. These timings are only as reference, as they will differ with each application, depending on the setup of the IQS263. For example, the sense (or charge time) is affected by the target counts and charge transfer frequency, while process time is dependent on the turbo mode activation, ATI checking for counts within the allowed band, filter settings and slider calculations. Communication time is affected by the MCU clock speed and the amount of data read (as well as the sequence thereof). Communication time and the number of active channels will influence the Scan Period.

### 8.2.2 Low Power rate

A wide range of low current consumption charging modes is available on the IQS263.

In any Low Power (LP) mode, there will be an applicable low power time (tLP).

With the detection of an undebounced proximity event the IC will zoom to BP mode, allowing a very fast reaction time for further possible touch events.

[^1]During any LP mode, only channel 0 is charged every tLp. The LP charge timing is illustrated in Figure 6.1.

If a low power rate is selected and charging is not in the zoomed state ( BP mode), the low power active bit (Register 0x01) will be set.

## Please refer to Section 6.12.

### 8.3 Touch report Rate

During Boost Power (BP) mode, the touch report rate of the IQS263B device depends on the charge transfer frequency, the number of channels enabled and the length of communications performed by the host MCU or master device.

Several factors may influence the touch report rate:

- Enabled channels: Disabling channels that are not used will not only increase the touch report rate, but will also reduce the device's current consumption.
- Turbo Mode: See Section 6.5.6
- Target Values: Lower target values requires shorter charge transfer times (tsense), thus reducing the SCAN PERIOD and increasing the touch report rate.
- Charge Transfer Speed: Increasing the charge transfer frequency will reduce the conversion time (tsense) and increase the touch report rate.
- ACF: Disabling the AC filter and wheel/slider position calculations will reduce the processing time (tprocess) and yield a faster report rate.


### 8.4 Long Term Average

The Long-term Average (LTA) filter can be seen as the baseline or reference value. The LTA is calculated to continuously adapt to
any environmental drift. The LTA filter is calculated from the CS value for each channel. The LTA filter allows the device to adapt to environmental (slow moving) changes/drift. Actuation (Touch or Prox) decisions are made by comparing the CS value with the LTA reference value.

### 8.5 Determine Touch or Prox

An event is determined by comparing the CS value with the LTA. Since the CS reacts differently when comparing the self- with the mutual capacitance technology, the user should consider only the conditions for the technology used.

An event is recorded if:

- Self: CS < LTA - Threshold
- Projected: CS > LTA + Threshold

Threshold can be either a Proximity or Touch threshold, depending on the current channel being processed.

Note that a proximity condition will be forced enabled if there is a touch condition on any channel.

Please refer to Section 6.7 and 6.8 for proximity and touch threshold selections.

### 8.6 ATI

The Automatic Tuning Implementation (ATI) is a sophisticated technology implemented on the new ProxSense ${ }^{\circledR}$ series devices. It allows for optimal performance of the devices for a wide range of sense electrode capacitances, without modification or addition of external components.

The ATI allows the tuning of two parameters, an ATI Multiplier and an ATI Compensation, to adjust the Count values for an attached sense electrode.

ATI allows the designer to optimize a specific design by adjusting the sensitivity
and stability of each channel through the adjustment of the ATI parameters.

The IQS263B has a full ATI function. The full-ATI function is default enabled, but can be disabled by setting the ATI_OFF bit, or changed to partial or alternative ATI by setting the ATI_Partial and ATI_ALT.
The ATI_Busy bit will be set while an ATI event is busy.

For more information regarding the ATI algorithm, please contact Azoteq at: ProxSenseSupport@azoteq.com

### 8.6.1 ATI Sensitivity

On the IQS263B device, the user can specify the BASE value (Section 6.13) and the TARGET value (Section 6.14) for the proximity channel ( CHO ) and touch channels (CH1-CH3).

A rough estimation of sensitivity can be calculated as:

$$
\text { Sensitivity } \propto \frac{T A R G E T}{B A S E}
$$

As can be seen from this equation, the sensitivity can be increased by either increasing the Target value or decreasing the Base value. It should, however, be noted that a higher sensitivity will yield a higher noise susceptibility.

### 8.6.2 ATI Target

The target value is reached by adjusting the COMPENSATION bits for each channel (ATI target limited to 2048 counts).

The target value is written into the respective channel's TARGET registers. The value written into these registers multiplied by 8 will yield the new target value. (Please refer to Section 6.14)

### 8.6.3 ATI Base (Multiplier)

The following parameters will influence the base value:

- $\mathrm{C}_{\mathrm{s}}$ SIZE ${ }^{1}$ : Size of sampling capacitor.
- PROJ_BIAS bits: Adjusts the biasing of some analogue parameters in the mutual capacitive operated IC. (Only applicable in mutual capacitance mode.)
- Charge Transfer Frequency
- MULTIPLIER bits.

The base value used for the ATI function can be implemented in 2 ways:

1. ATI_PARTIAL $=0$. ATI automatically adjusts MULTIPLIER bits to reach a selected base value ${ }^{2}$. Please refer to Section 6.13 for available base values.
2. ATI_PARTIAL = 1. The designer can specify the multiplier settings. These settings will give a custom base value from where the compensation bits will be automatically implemented to reach the required target value. The base value is determined by two sets of multiplier bits. Sensitivity Multipliers which will also scale the compensation to normalise the sensitivity and Compensation Multipliers to adjust the gain.

### 8.6.4 Re-ATI

An automatic re-ATl event will occur if the counts are outside its re-ATI limits. The reATI limit or ATI boundary is calculated as the target value divided by 8 . For example:

- Target $=512$, Re-ATI will occur if CS is outside 512 $\pm 64$.

[^2][^3]IQ Switch ${ }^{\circledR}$
ProxSense ${ }^{\circledR}$ Series
8.6.5 ATI ERROR

The ATI error bit (read only) indicates to the user that the ATI targets where not reached. Adjustments of the base values or ATI BANDs are required.

## 9 Communication

The IQS263B device interfaces to a master controller via a 3-wire (SDA, SCL and RDY) serial interface bus that is $I^{2} C^{\top M}$ compatible, with a maximum communication speed of $400 \mathrm{kbit} / \mathrm{s}$.

### 9.1 Control Byte

The Control byte indicates the 7-bit device address (44H default) and the Read/Write indicator bit. The structure of the control byte is shown in Figure 9.1.


Figure 9.1 IQS263B Control Byte.
The $I^{2} \mathrm{C}$ device has a 7 bit Slave Address (default 0 x 44 H ) in the control byte as shown in Figure 9.1. To confirm the address, the software compares the received address with the device address. Sub-address values can be set by OTP programming options.

## $9.2 \mathrm{I}^{2} \mathrm{C}$ Read

To read from the device a current address read can be performed. This assumes that the address-command is already setup as desired.

Current Address Read

| Start | Control Byte | Data n | Data $\mathrm{n}+1$ | Stop |
| :---: | :---: | :---: | :---: | :---: |
| S | ACK | ACK | NACK | S |

## Figure 9.2 Current Address Read.

If the address-command must first be specified, then a random read must be performed. In this case a WRITE is initially performed to setup the address-command, and then a repeated start is used to initiate the READ section.

|  | Random Read |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Start | Control Byte |  |  | Star | Control Byte |  |  | Stop |
| S | Adr + WRITE | ACK | ACK | S | Adr + READ | ACK | NACK | s |

Figure 9.3 Random Read.

### 9.3 I $^{2} \mathrm{C}$ Write

To write settings to the device a Data Write is performed. Here the Address-Command is always required, followed by the relevant data bytes to write to the device.


Figure $9.4 \quad I^{2} \mathrm{C}$ Write.

### 9.4 End of Communication Session / Window

Similar to other Azoteq ${ }^{2} \mathrm{C}$ devices, to end the $\mathrm{I}^{2} \mathrm{C}$ communication session, a STOP command is given. When sending numerous read and write commands in one communication cycle, a repeated start command must be used to stack them together (since a STOP will jump out of the communication window, which is not desired).
The STOP will then end the communication, and the IQS263B will return to process a new set of data. Once this is obtained, the communication window will again become available (RDY set LOW).

## $9.5 \mathrm{I}^{2} \mathrm{C}$ Sub-address

The IQS263B has four available sub addresses, 44 H (default) to 47 H , which allows up to four devices on a single $\mathrm{I}^{2} \mathrm{C}$ bus.

### 9.5.1 Internal sub-address selection

Selecting the sub-address via OTP bits allows the user 4 different options:
Table $9.1 \quad I^{2} \mathrm{C}$ sub-address selection

| FG25 | FG26 | Device Address |
| :---: | :---: | :---: |
| 0 | 0 | $0 \times 44$ |
| 0 | 1 | $0 \times 45$ |
| 1 | 0 | $0 \times 46$ |
| 1 | 1 | $0 \times 47$ |

### 9.6 RDY Hand-Shake Routine

The master or host MCU has the capability to request a communication window at any time, by pulling the RDY line low. The communication window will open directly following the current conversion cycle. For more details please refer to the communication interface guide.

### 9.7 I $^{2} \mathrm{C}$ Specific Commands

### 9.7.1 Show Reset

After start-up, and after every reset event, the "Show Reset" flag will be set in the System Flags register (0x01H; byte 0).
The "Show Reset" bit can be read to determine whether a reset has occurred on the device (it is recommended to be continuously monitored). This bit will be set '1' after a reset.

The SHOW_RESET bit will be cleared (set to ' 0 ') by writing a ' 0 ' into the "Show Reset"
bit. A reset will typically take place if a timeout during communication occurs.

### 9.7.2 I2C Timeout

If no communication is initiated from the master/host MCU within the first $\mathrm{t}_{\text {comms }}$ (tсоммs $=5.12 \mathrm{~ms}$ default) of the RDY line indicating that data is available (i.e. RDY = low), the device will resume with the next cycle of charge transfers and the data from the previous conversions will be lost. The timeout time is adjustable in steps of 1.28 ms in the Thresholds register ( $0 \times 0 \mathrm{~A}$; byte 7). There is also a timeout ( $\mathrm{t}_{12 \mathrm{C}}$ ) that cannot be disabled, for when communication has started but not been completed, for example when the bus is being held by another device. $\mathrm{t}_{12 \mathrm{C}}=150 \mathrm{~ms}$.

### 9.8 I $^{2} \mathrm{C}$ I/O Characteristics

The IQS263B requires the input voltages given in Table 9.2, for detecting high ("1") and low (" 0 ") input conditions on the $I^{2} \mathrm{C}$ communication lines (SDA, SCL and RDY).

Table 9.2 IQS263B I ${ }^{2}$ C Input voltage

|  | Input Voltage (V) |
| :---: | :---: |
| VinLow | $0.3^{*} \mathrm{VDDHI}$ |
| Vin |  |

Table 9.3 provides the output voltage levels of the IQS263B device during $I^{2} \mathrm{C}$ communication.

Table 9.3 IQS263B I²C Output voltage

|  | Output Voltage (V) |
| :---: | :---: |
| Vout_ow | GND +0.2 (max.) |
| VoutHIGH | VDDHI -0.2 (min.) |

### 9.9 Multi -Slave Bus Operation

The IQS263B can operate on an $1^{2} \mathrm{C}$ bus with other slave devices. However, it should be noted that the IQS263B will clock stretch
between bits 8 and 9 when any device on the bus is addressed in order to determine whether the clocked address matches its own. This will only occur during times when the IQS263B is monitoring the $\mathrm{I}^{2} \mathrm{C}$ bus, this is whenever the ready line is low in both streaming and event mode. In event mode, there are also times during the operating cycle where the IQS263B will monitor the $I^{2} \mathrm{C}$ bus, regardless of whether an event has occurred, said clock stretching will also take place during these times. It is recommended to place any slave devices which cannot tolerate such behavior on a separate bus.

## 10 Communication Command/Address Structure

### 10.1 Registers \& Memory map

Table 10.1 IQS263B Registers

| Address | Description | Access | Section |
| :---: | :---: | :---: | :---: |
| $0 \times 00 \mathrm{H}$ | Device Information | R | 10.2 .1 |
| $0 \times 01 \mathrm{H}$ | System Flags | $\mathrm{R} / \mathrm{W}$ | 10.2 .2 |
| $0 \times 02 \mathrm{H}$ | Coordinates | R | 10.2 .3 |
| $0 \times 03 \mathrm{H}$ | Touch Bytes | R | 10.2 .4 |
| $0 \times 04 \mathrm{H}$ | Counts | R | 10.2 .5 |
| $0 \times 05 \mathrm{H}$ | LTA | R | 10.2 .6 |
| $0 \times 06 \mathrm{H}$ | Deltas | R | 10.2 .7 |
| $0 \times 07 \mathrm{H}$ | Multipliers | 10.2 .8 |  |
| $0 \times 08 \mathrm{H}$ | Compensation | $\mathrm{R} / \mathrm{W}$ | 10.2 .9 |
| $0 \times 09 \mathrm{H}$ | ProxSettings | $\mathrm{R} / \mathrm{W}$ | 10.2 .10 |
| $0 \times 0 \mathrm{AH}$ | Thresholds | $\mathrm{R} / \mathrm{W}$ | 10.2 .11 |
| $0 \times 0 \mathrm{BH}$ | $\mathrm{R} / \mathrm{W}$ | 10.2 .12 |  |

IQ Switch ${ }^{\circledR}$
ProxSense ${ }^{\circledR}$ Series

| $0 \times 0 \mathrm{CH}$ | Gesture Timers | $R / W$ | 10.2 .13 |
| :---: | :---: | :---: | :---: |
| $0 \times 0 \mathrm{DH}$ | Active Channels | $R / W$ | 10.2 .14 |

### 10.2 Registers Descriptions

### 10.2.1 Device Information 0x00H

Information regarding the device type and version is recorded here. Any other information specific to the device version can be stored here. Each Azoteq ROM has a unique Product- and Version number.

|  |  | Product Number (PROD_NUM) |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R | Value | 0x3C |  |  |  |  |  |  |  |


|  | Version Number (VERSION_NUM) |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access |  |  |  |  |  |  |  |  |
| R |  |  |  |  |  |  |  |  | | Bit | 7 | 6 | 5 | 4 | 3 | 2 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |

10.2.2 System Flags 0x01H

System Flags (SYSFLAGSO)

| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| R/W | Name | $\begin{aligned} & \text { Show } \\ & \text { Reset } \end{aligned}$ | Move ment | $\begin{aligned} & \frac{\text { ATI }}{\text { Error }} \end{aligned}$ | Proj <br> Mode | Filter <br> Halt | ATI Busy | Ind <br> Halt | LP <br> Active |

Bit 7: $\quad 0=$ Reset has NOT occurred.
1 = Reset has occurred. Write to 0 to clear.
Bit 6: $\quad 0=$ No movement detected.
1 = Movement detected.
Bit 5: $\quad 0=$ ATI Error has NOT occurred.
1 = ATI Error has occurred.
Bit 4: $\quad 0=$ Self capacitive mode selected.

1 = Projected capacitive mode selected.
Bit 3: Will show only if "Ind Halt" bit is set to 0 . $0=A$ global filter halt is not active. 1 = A global filter halt is active.
Bit 2: $\quad 0=$ Device is not busy with ATI. 1 = Device is busy with ATI.
Bit 1: $\quad 0=$ Global Filter Halt enabled. 1 = Global Filter Halt disabled.

Bit 0: $\quad 0=$ Low Power mode is not active. 1 = Low Power mode is active.

|  | Events |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Access | Bit | $\mathbf{7}$ | $\mathbf{6}$ | $\mathbf{5}$ | $\mathbf{4}$ | $\mathbf{3}$ | $\mathbf{2}$ | $\mathbf{1}$ |
| $\mathbf{R}$ | Flick <br> Right | Flick <br> Left | Tap | Move <br> ment | ATI <br> Event | Slide <br> Event | Touch <br> Event | Prox <br> Event |  |

### 10.2.3 Wheel Coordinates 0x02H

|  |  | Wheel 1 Low |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R | Name | Wheel 1 Coordinate Low byte first |  |  |  |  |  |  |  |


|  |  | Relative Coordinate Low |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R | Name | Relative Coordinate Low byte first |  |  |  |  |  |  |  |


|  |  | Relative Coordinate High |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R | Name | Relative coordinate High byte |  |  |  |  |  |  |  |

10.2.4 Touch Bytes 0x03H

|  |  | Touch Byte 0 |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R | Name |  |  |  |  | CH3 | CH2 | CH1 | $\mathrm{CH} 0^{1}$ |


|  |  | Halt Byte 1 |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R | Name |  |  |  |  | CH3 | CH2 | CH 1 | $\mathrm{CHO}{ }^{2}$ |

Bit 0 of the first byte ( CH 0 ) will indicate proximity events; the rest of the bits indicate touches as shown. The second byte shows the halt status bits.

### 10.2.5 Counts 0x04H

This register has 10 bytes to store the count values of the low power channel and then CHO up to CH 3 the low byte will always read out first, followed by the high byte, before the moving to the next channel.

|  | Low Power Channel Low |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| $\mathbf{R}$ |  |  |  |  |  |  |  |  |  |
| Name |  |  |  |  |  |  |  |  |  |
| Byte 0 |  |  |  |  |  |  |  |  |  |


|  |  | CH 3 Counts High byte |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R | Name | Channel 3 Count value (High byte last) |  |  |  |  |  |  |  |
| Byte 9 |  |  |  |  |  |  |  |  |  |

### 10.2.6 LTA 0x05H

This register has 10 bytes to store the LTA values of the low power channel and then CHO up to CH 3 the low byte will always read out first, followed by the high byte, before the moving to the next channel.
${ }^{1} \mathrm{CHO}$ indicates Proximity, not Touch.
${ }^{2} \mathrm{CH} 0$ indicates Proximity, not Touch.

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|  | Low Power Channel LTA Low byte |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R |  |  |  |  |  |  |  |  |  |
| Name |  |  |  |  |  |  |  |  |  |
| Byte 0 |  |  |  |  |  |  |  |  |  |


|  |  | CH 3 LTA High byte |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R | Name | Channel 3, LTA value (High byte last) |  |  |  |  |  |  |  |
| Byte 9 |  |  |  |  |  |  |  |  |  |

### 10.2.7 Deltas 0x06H

This register has 8 bytes to store the Delta (the difference between Count and LTA) values of CHO up to CH 3 . The low byte will always read out first, followed by the high byte, before the moving to the next channel. Deltas are not available when using the wheel setting.

|  | Delta Counts for CHO Low |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access |  |  |  |  |  |  |  |  |
| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| Byte 0 |  |  |  |  |  |  |  |  |
| Name |  |  |  |  |  |  |  |  |
|  |  | Delta for CHO - Low Byte first |  |  |  |  |  |  |


|  |  | Delta Counts for CH3 High |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R | Name | Delta for CH3 - High Byte last |  |  |  |  |  |  |  |
| Byte 7 |  |  |  |  |  |  |  |  |  |

### 10.2.8 Multipliers 0x07H

|  |  | CHO Multipliers |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W | Name | Sensitivity Multipliers |  | Sensitivity <br> Multipliers |  | Comp Multipliers |  |  |  |

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Byte 0

|  |  |
| :--- | :--- |

$\square \mid \square$

|  | CH1 Multipliers |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W |  |  | Sensitivity <br> Multipliers | Comp Multipliers |  |  |  |  |  |
| Name |  |  |  |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  |  |



|  |  | CH1/CH0 Base Values |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W | Name | Channel 1 |  |  |  | Channel 0 |  |  |  |
| Byte 4 | Default | $0 \times 44$ |  |  |  |  |  |  |  |


|  | $\mathrm{CH} / \mathrm{CH} 2$ Base Values |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W |  |  |  |  |  |  |  |  |  |
| Byte 5 |  |  |  |  |  |  |  |  |  |
| Name | Channel 3 |  |  |  |  |  |  |  |  |
| Default | Channel 2 |  |  |  |  |  |  |  |  |

Base Value Options: IQ Switch ${ }^{\circledR}$ ProxSense ${ }^{\circledR}$ Series

| 0000 | -74 | 1000 | -202 |
| :--- | :--- | :--- | :--- |
| 0001 | -90 | 1001 | -218 |
| 0010 | -106 | 1010 | -234 |
| 0011 | -122 | 1011 | -250 |
| 0100 | -138 (default) | 1100 | -266 |
| 0101 | -154 | 1101 | -282 |
| 0110 | -170 | 1110 | -298 |
| 0111 | -186 | 1111 | -314 |

### 10.2.9 Compensation 0x08H

|  |  | CHO Compensation value |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W | Name | Channel 0 Compensation |  |  |  |  |  |  |  |
| Byte 0 |  |  |  |  |  |  |  |  |  |


|  |  | CH 3 Compensation Value |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W | Name | Last active channel Compensation |  |  |  |  |  |  |  |
| Byte 3 |  |  |  |  |  |  |  |  |  |

10.2.10 ProxSettings 0x09H

|  |  | ProxSettings0 |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W | Name | $\frac{\underline{\text { ATI }}}{\underline{\text { OFF }}}$ | ATI <br> Partial | ATI <br> Band | $\begin{aligned} & \underline{\text { Redo }} \\ & \underline{\text { ATI }} \end{aligned}$ | Reseed | Stream <br> ATI | $\underline{4 M H z}$ | Force <br> Halt |


| Byte 0 | Default | OxOOH |
| :---: | :---: | :---: |
| Bit 7: | $\begin{aligned} & 0=\text { ATI } \\ & 1=\text { ATI } \end{aligned}$ | Enable Disabled |
| Bit 6: | $\begin{aligned} & 0=\text { Full } A \\ & 1=\text { Partia } \end{aligned}$ | ATI <br> ial ATI |
| Bit 5: | $\begin{aligned} & 0=1 / 8 * \\ & 1=1 / 4 * \end{aligned}$ |  |
| Bit 4: | $\begin{aligned} & 0=\text { No A } \\ & 1=\text { Redo } \end{aligned}$ |  |
| Bit 3: | $\begin{aligned} & 0=\text { No } A \\ & 1=\text { Send } \end{aligned}$ | Action <br> d Reseed |
| Bit 2: | $\begin{aligned} & 0=\text { No cc } \\ & 1=\text { Comr } \end{aligned}$ | communication widows during ATI routine munication windows after each charge cycle regardless of ATI busy. |
| Bit 1: | $\begin{aligned} & 0=2 \mathrm{MH}_{2} \\ & 1=4 \mathrm{MH}_{2} \end{aligned}$ | Hz Oscillator Hz Oscillator |
| Bit 0: | $\begin{aligned} & 0=\text { No A } \\ & 1=\text { Forci } \end{aligned}$ | Action <br> cing all LTAs to stop calculating |

ProxSettings1

| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| R/W | Name | $\frac{\text { WDT }}{\underline{\text { OFF }}}$ | Event <br> Mode |  |  |  |  |  |  |
| Byte 1 | Default | $0 \times 11 \mathrm{H}$ |  |  |  |  |  |  |  |

Bit 7: $\quad 0=$ WDT Enable
1 = WDT Disabled
Bit 6: $\quad 0=$ Streaming Mode
1 = Event Mode
Bit 5:4: $\quad 00=2^{9}$
$01=2^{8}$ (default)
$10=2^{7}$
$11=2^{6}$ (fastest following)
Bit 3:2: $\quad 00=$ Slider Disabled
$01=2 \mathrm{CH}$ Slider
$10=$ Wheel (Also used for 3CH wrap around slider)
$11=3 \mathrm{CH}$ Slider
Bit 1:0: $\quad 00=$ Counts Filtering OFF
$01=$ Beta -01
$10=$ Beta -02
11 = Beta - 03 (largest filter for noise suppression, slowest response)

|  |  | ProxSettings2 |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W | Name | $\begin{aligned} & \text { Sleep } \\ & \underline{\text { Halt }} \end{aligned}$ | Force Sleep | Wake <br> Release | Wheel <br> Filter | Mov | ent |  |  |
| Byte 2 | Default | OxOOH |  |  |  |  |  |  |  |

Bit 7: $\quad 0=$ Reseed upon Halt timer expiration
1 = Return to LP upon Halt timer expiration
Bit 6:
$0=$ No Action
1 = Send Force Sleep
Bit 5: $\quad 0=$ Wake from LP only in the normal direction (Counts below LTA for Self)
$1=$ Wake from LP for counts in both direction
Bit 4: $\quad 0=$ Coordinate Filter Enabled
1 = Coordinate Filter Disabled
Bit 3:2: $\quad 00=$ Movement Disabled
$01=$ Movement on CH 0
$10=$ Movement on CH3
Bit 1:0:
00 = Prox
$01=$ Sync (ZC input)
$10=$ Touch on CH1
11 = Movement output

|  |  | ProxSettings 3 |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W | Name | Touch <br> Debounc <br> e | Off <br> Mode | Proj <br> Bias | Float <br> Cx | Halt <br> Charge | Turbo <br> Mode | Filtered <br> Touches | $\underline{\text { Xfer }}$ |
| Byte 3 | Default | OxOOH |  |  |  |  |  |  |  |

Bit 7: $\quad 0=2$ consecutive samples

Bit 6: $\quad 0=$ off mode disabled
1 = off mode enabled
Bit 5: $\quad 0=10 \mathrm{uA}$
$1=20 u A$
Bit 4: $\quad 0=$ Grounded
1 = Floating
Bit 3: $\quad 0=$ Normal Charging

1 = Halt All Channels' Charging
Bit 2:
0 = Turbo Mode Enabled
1 = Turbo Mode Disabled
Bit 1: $\quad 0=$ Touch detection on unfiltered counts
$1=$ Touch detection on filtered counts
Bit 0:
0 = Fast Charging
1 = Slower Charging

|  |  | Event Mask |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W | Name | Flick Left | Flick Right | Tap | Move ment | ATI | Slide | Touch | Prox |
| Byte 4 | Default | OxFFH |  |  |  |  |  |  |  |

10.2.11 Thresholds 0x0A

Proximity Threshold

| Proximity Threshold |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W |  |  |  |  |  |  |  |  |  |
| Byte 0 |  |  |  |  |  |  |  |  |  |
| Name | Value |  |  |  |  |  |  |  |  |
| Default | 4 C |  |  |  |  |  |  |  |  |


|  |  | Touch Threshold CH1 |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W | Name |  |  |  |  |  |  |  |  |
| Byte 1 | Default |  |  |  |  |  |  |  |  |

Touch Threshold CH3

| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| R/W |  |  |  |  |  |  |  |  |  |
| Byte 3 |  |  |  |  |  |  |  |  |  |
| Name | $1-255$ |  |  |  |  |  |  |  |  |
| Default | 16 D |  |  |  |  |  |  |  |  |

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$\int$ Azoteq

|  |  | Movement Threshold |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W | Name |  |  |  |  |  |  |  |  |
| Byte 4 | Default |  |  |  |  |  |  |  |  |


|  |  | Halt timeout Reseed Block |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W | Name |  |  |  |  | CH3 | CH 2 | CH 1 |  |
| Byte 5 | Default |  |  |  |  |  |  |  |  |


|  | Halt Time in Number of Samples |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access |  |  |  |  |  |  |  |  |
| R/W |  |  |  |  |  |  |  |  |
| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| Byte 6 |  |  |  |  |  |  |  |  |
| Name |  |  |  |  |  |  |  |  |
| Default | Value times 50 (FF = Always; 00 = Never) |  |  |  |  |  |  |  |


|  |  | I2C Timeout |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W | Name | Value times 1.28 ms |  |  |  |  |  |  |  |
| Byte 7 | Default | 4D |  |  |  |  |  |  |  |

10.2.12 Timings $0 \times 0 B H$

|  |  | Low Power Time |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W | Name | Steps of 16 ms (Value times 16 ms ) |  |  |  |  |  |  |  |
| Byte 0 | Default | OxOOH |  |  |  |  |  |  |  | IQ Switch ${ }^{\circledR}$ ProxSense ${ }^{\circledR}$ Series


|  | ATI Target for Touch Channels |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W |  |  |  |  |  |  |  |  |  |
| Byte 1 |  |  |  |  |  |  |  |  |  |
| Name |  |  |  |  |  |  |  |  |  |
| Default | Steps of 8 (Value times 8) |  |  |  |  |  |  |  |  |


|  | ATI Target for Proximity |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W |  |  |  |  |  |  |  |  |  |
| Byte 2 |  |  |  |  |  |  |  |  |  |
| Name |  |  |  |  |  |  |  |  |  |
| Default | Steps of 8 (Value times 8) |  |  |  |  |  |  |  |  |

10.2.13 Gesture Timers $0 \times 0 \mathrm{CH}$

|  |  | Tap Timer |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W | Name | Tap Timer Limit |  |  |  |  |  |  |  |
| Byte 0 | Default | 5D |  |  |  |  |  |  |  |


|  | Flick Timer |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W |  |  |  |  |  |  |  |  |  |
| Byte 1 |  |  |  |  |  |  |  |  |  |
| Name | Flick Timer Limit |  |  |  |  |  |  |  |  |
| Default |  |  |  |  |  |  |  |  |  |


|  | Flick Threshold |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access |  |  |  |  |  |  |  |  |
| R/W |  |  |  |  |  |  |  |  |
| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| Byte 2 |  |  |  |  |  |  |  |  |
| Nefault |  |  |  |  |  |  |  |  |
| Flick Threshold Value |  |  |  |  |  |  |  |  |

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10.2.14 Active Channels 0x0DH

|  |  | Active Chan 0 |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Access | Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| R/W | Name |  |  |  |  | CH3 | CH 2 | CH1 | CHO |
| Byte 0 | Default | 0x0FH |  |  |  |  |  |  |  |

## 11 Specifications

### 11.1 Absolute Maximum Specifications

The following absolute maximum parameters are specified for the device:
Exceeding these maximum specifications may cause damage to the device.

- Operating temperature
- Supply Voltage (VDDHI - GND)
- Maximum pin voltage
- Maximum continuous current (for specific Pins)
- Minimum pin voltage
- Minimum power-on slope
- ESD protection
$-20^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$
3.6 V

VDDHI +0.5 V (may not exceed VDDHI max)
10 mA
GND - 0.5V
100V/s
$\pm 4 \mathrm{kV}$ (Human body model)

Table 11.1 IQS263B Self Capacitive General Operating Conditions ${ }^{1}$

| DESCRIPTION | Conditions | PARAMETER | MIN | TYP | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Supply voltage |  | $V_{\text {DDHI }}$ | 1.8 | 3.3 V | 3.6 | V |
| Internal regulator output | $1.8 \leq \mathrm{V}_{\text {DDHI }} \leq 3.6$ | $\mathrm{~V}_{\text {REG }}$ | 1.62 | 1.7 | 1.79 | V |
| Streaming mode | 3.3 V | 80 Hz |  | 180 |  | $\mu \mathrm{~A}$ |
| Event Mode |  | 80 Hz |  | 90 | 150 | $\mu \mathrm{~A}$ |
| ${\text { Low Power Setting } 8^{* * *}}$ |  | 128 ms |  | 4 | 6.5 | $\mu \mathrm{~A}$ |
| Low Power Setting $16^{* *}$ |  | 256 ms |  | 3 | 4 | $\mu \mathrm{~A}$ |

Table 11.2 IQS263B Projected Capacitive General Operating Conditions

| DESCRIPTION | Conditions | PARAMETER | MIN | TYP | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Supply voltage |  | $V_{\text {DDHI }}$ | 1.8 | 3.3 V | 3.6 | V |
| Internal regulator output | $1.8 \leq \mathrm{V}_{\text {DDHI }} \leq 3.6$ | $\mathrm{~V}_{\text {REG }}$ | 1.62 | 1.7 | 1.79 | V |
| Streaming mode | 3.3 V | 80 Hz | - | 305 |  | $\mu \mathrm{~A}$ |
| Event Mode |  | 80 Hz | - | 230 |  | $\mu \mathrm{~A}$ |
| Low Power Setting $8^{* *}$ |  | 128 ms | - | 5 | 11 | $\mu \mathrm{~A}$ |
| Low Power Setting 16** |  | 256 ms | - | 4 | 6 | $\mu \mathrm{~A}$ |

[^4]IQ Switch ${ }^{\circledR}$
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"LP interval period = Low power value $\times 16 \mathrm{~ms}$
Table 11.3 Start-up and shut-down slope Characteristics

| DESCRIPTION | Conditions | PARAMETER | MIN | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Power On Reset | $V_{\text {DDHI }}$ Slope $\geq 100 \mathrm{~V} / \mathrm{s}$ <br> $@ 25^{\circ} \mathrm{C}$ | POR |  | 1.6 | V |
| Brown Out Detect | $\mathrm{V}_{\text {DDHI }}$ Slope $\geq 100 \mathrm{~V} / \mathrm{s}$ <br> $@ 25^{\circ} \mathrm{C}$ | BOD | 1.05 |  | V |

Table 11.4 Electrode Specifications - Self Capacitance

| DESCRIPTION | Conditions | PARAMETER | MAX | UNIT |
| :--- | :--- | :---: | :---: | :---: |
| Parasitic Capacitance CX to GND |  | $C_{P}$ | 120 | pF |
| Series Resistor | $\mathrm{C}_{\mathrm{P}}=120 \mathrm{pF}$ | $\mathrm{R}_{\mathrm{S}}$ | 10 | $\mathrm{k} \Omega$ |

Table 11.5 Electrode Specifications - Mutual Capacitance

| DESCRIPTION | Conditions | PARAMETER | MIN | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Parasitic Capacitance Tx to GND |  | $C_{T}$ |  | 100 | pF |
| Parasitic Capacitance Rx to GND |  | $\mathrm{C}_{\mathrm{R}}$ |  | 100 | pF |
| Mutual Capacitance |  | $\mathrm{C}_{\mathrm{M}}$ | 0.1 | 10 | pF |
| Series Resistor |  | $\mathrm{R}_{\mathrm{TX}}$ |  | 10 | $\mathrm{k} \Omega$ |
| Series Resistor | $\mathrm{C}_{\mathrm{M}}=1 \mathrm{pF}$ | $\mathrm{R}_{\mathrm{RX}}$ |  | 1 | $\mathrm{k} \Omega$ |

IQ Switch ${ }^{\circledR}$
ProxSense ${ }^{\circledR}$ Series
Table 11.6 ATI Times

| Turbo Mode Off |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Oscillator | 2MHZ |  |  |  | 4MHZ |  |  |  |
| Target value | Low |  | High |  | Low |  | High |  |
| Channels active | 1 | 3 | 1 | 3 | 1 | 3 | 1 | 3 |
| Typical time [ms] | 625 | 625 | 630 | 630 | 310 | 300 | 313 | 305 |
| Worst case | 850 (34 cycles @ 25 ms per cycle) |  |  |  |  |  |  |  |
| Turbo Mode On |  |  |  |  |  |  |  |  |
| Oscillator | 2MHZ |  |  |  | 4MHZ |  |  |  |
| Target value | Low |  | High |  | Low |  | High |  |
| Channels active | 1 | 3 | 1 | 3 | 1 | 3 | 1 | 3 |
| Typical time [ms] | 200 | 350 | 240 | 420 | 105 | 175 | 120 | 205 |
| Worst case | 500-600 (34 cycles estimate) |  |  |  |  |  |  |  |
| ATI Error / Failure |  |  |  |  |  |  |  |  |
| Worst case Including re-tries |  | 4.7 seconds ( 189 cycles <br> @25ms per cycle) |  |  | Something is wrong with settings or electrode(s) |  |  |  |

## 12 Packaging Information

### 12.1 MSOP-10



## NOTES:

1) THIS DRAWING IS FOR CUSTOMER PURPOSE.
2) LEADFRAME MATERIAL : C7025(THICKNESS: 0.15 MM ).
3) BOTH PACKAGE LENGTH \& WIDTH DO NOT INCLUDE MOLD FLASH.
4) FORMED LEAD SHALL BE PLANAR WITH RESPECT TO ONE ANOTHER WITHIN 0.10 MM .
5) UNREMOVED FLASH BETWEEN LEADS \& PACKAGE END FLASH SHALL NOT EXCEED 0.15 MM FROM BOTTOM BODY.
6) CONTROLLING DIMENSION : MM(INCH).

Figure 12.1 MSOP-10 Package Dimensions.


Figure 12.2 MSOP-10 Footprint.

Table 12.1 MSOP-10 Footprint Dimensions from Figure 12.2.

| Dimension | [mm] |
| :---: | :---: |
| Pitch | 0.50 |
| C | 4.40 |
| Y | 1.45 |
| $X$ | 0.30 |

12.2 DFN10

Table 12.2 DFN-10 Package dimensions (bottom)

| Dimension | $[\mathrm{mm}]$ |
| :---: | :---: |
| A | $3 \pm 0.1$ |
| B | 0.5 |
| C | 0.25 |
| D | n/a |
| F | $3 \pm 0.1$ |
| L | 0.4 |
| P | 2.4 |
| Q | 1.65 |



Figure 12.3 DFN-10 Package dimensions (bottom). Note that the saddle need to be connected to GND on the PCB.
Table 12.3 DFN-10 Package dimensions (side)

| Dimension | $[\mathrm{mm}]$ |
| :---: | :---: |
| G | 0.05 |
| H | 0.65 |
| I | $0.7-0.8$ |

Table 12.4 DFN-10 Landing dimensions

| Dimension | $[\mathrm{mm}]$ |
| :---: | :---: |
| A | 2.4 |
| B | 1.65 |
| C | 0.8 |
| D | 0.5 |
| E | 0.3 |
| F | 3.2 |



Figure 12.4 DFN-10 Package dimensions (side)


Figure 12.5DFN-10 Landing dimensions
12.3Tape and Reel Specification
12.3.1 MSOP10


NOTES:

1. 10 sprocket hole pitch caumulative tolerance $\pm 0.2$
2. Camber not to exceed 1 mm in 100 mm
3. Material:Black conductive Polystyrene
4. Ao and Bo measured on a plane 0.3 mm above the bottom of the pocket
5. Ko measured from a plane on the inside bottom of the pocket to the top surface of the carrier
6. Pocket position relative to sprocket hole measured as

true position of pocket, not pocket hole.
7. Pocket center and pocket hole center must be same position

Figure 12.6 MSOP-10 Tape Specification. Bulk orientation LT.
12.3.2 DFN10 (3x3)


SECTIDN A - 4
$A 0=3.30$
$B 0=3.30$
$K 0=1.10$
NOTES:

1. 10 SPROCKET HOLE PITCH CUMULATVE TOLERANCE $\pm 0.2$
2. CAMBER IN COMPLAACE WTH EIA 481
3. POCKET POSITION RELATVE TO SPROCKET HOLE

MEASURED AS TRUE POSITION OF POCKET, NOT POCKET HOLE


### 12.4Package MSL

Moisture Sensitivity Level (MSL) relates to the packaging and handling precautions for some semiconductors. The MSL is an electronic standard for the time period in which a moisture sensitive device can be exposed to ambient room conditions (approximately $30^{\circ} \mathrm{C} / 85 \% \mathrm{RH}$ see J-STD033C for more info) before reflow occur.

Table 12.5 MSL

| Package | Level (duration) |
| :--- | :--- |
| MSOP-10 | MSL 1 (Unlimited at $\leq 30^{\circ} \mathrm{C} / 85 \% \mathrm{RH}$ ) |
|  | Reflow profile peak temperature $<260^{\circ} \mathrm{C}$ for $<25$ seconds <br> Number of Reflow $\leq 3$ |
| DFN10 (3×3) | MSL 1 (Unlimited at $\left.\leq 30^{\circ} \mathrm{C} / 85 \% ~ R H\right)$ <br> Reflow profile peak temperature $<260^{\circ} \mathrm{C}$ for $<25$ seconds <br> Number of Reflow $\leq 3$ |

## 13 Device Marking

### 13.1 Top Marking



| REVISION | $\mathbf{x}$ | $=$ | IC Revision Number |
| :--- | :--- | :--- | :--- |
| TEMPERATURE RANGE | $\mathbf{t}$ | $=$ | i $-20^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$ (Industrial) |
|  |  | $=$ | $\mathrm{C} \quad 0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}$ (Commercial) |
| IC CONFIGURATION | $\mathbf{z}$ | $=$ | Configuration (Hexadecimal) |
|  |  |  | Marking not present for a configuration of 0. |
| DATE CODE | $\mathbf{P}$ | $=$ | Package House |
|  | $\mathbf{W W}$ | $=$ | Week |
|  | YY | $=$ | Year |
|  |  |  |  |

## 14 Ordering Information

Order quantities will be subject to multiples of a full reel. Contact the official distributor for sample quantities. A list of the distributors can be found under the "Distributors" section of www.azoteq.com.

### 14.1 MSOP-10 Package



| IC NAME | IQS263B | $=$ | IQS263B |
| :--- | :--- | :--- | :--- |
| CONFIGURATION | $z$ | $=$ | Sub Address Configuration (hexadecimal) |
|  |  | $0=44 \mathrm{H}$ |  |
|  |  | $1=45 \mathrm{H}$ |  |
|  |  | $2=46 \mathrm{H}$ |  |
|  |  | $3=47 \mathrm{H}$ |  |
| PACKAGE TYPE | MS | $=$ | MSOP-10 |
| BULK PACKAGING | $R$ | $=$ | Reel MSOP - 4000pcs/reel |

### 14.2 DFN Package



| IC NAME | IQS263B | $=$ | IQS263B |
| :--- | :--- | :--- | :--- |
| CONFIGURATION | $z$ | $=$ | Sub Address Configuration (hexadecimal) |
|  |  | $0=44 \mathrm{H}$ |  |
|  |  | $1=45 \mathrm{H}$ |  |
|  |  | $2=46 \mathrm{H}$ |  |
|  |  | $3=47 \mathrm{H}$ |  |
| PACKAGE TYPE | DN | $=$ | DFN10 $(3 \times 3)$ |
| BULK PACKAGING | R | $=$ | Reel DNF10 $(3 \times 3)-3000 \mathrm{pcs} /$ reel |


|  | USA | Asia |
| :--- | :---: | :---: | :---: |
| Physical |  |  |
| Address |  |  | | 11940 Jollyville |
| :---: |
| Suite 120-S |
| Austin |
| TX 78750 |
| USA |$\quad$| T-Share International Centre, |
| :---: |
| Taoyuan Road, Nanshan District, |
| Shenzhen, Guangdong, PRC |
| Address |

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CFAM 18P1600/S14 CFAM 30P1600 CQ28-10NPP-KW1 CR12CN04DNC KN056050 KN086050 KN186050 KV750450 KV750455 17-9291-226 98-0003-3048-4 CR12CN04DNO-E2 PROXISWITCH CR12CN04DPO BCS M12BBG1-PSC80H-EP02 BCS M30B4I1-PSC15DEP02 BCS M18B4G2-PSC15H-S04K CR18CF05DPO-E2 CM30-25NPP-EW1 CM30-25NNP-EW1 CM30-16BPP-EW1 CM30-16BNP-EW1 CM18-12NPP-EW1 CM18-12NPP-EC1 CM18-12NNP-EW1 CM18-12NNP-EC1 CM18-08BPP-EW1 CM18-08BNP-EW1 CM30-25NNPEC1


[^0]:    ${ }^{1}$ CS-LTA in Projected mode. LTA-CS in Self capacitive sensing mode.

[^1]:    ${ }^{1}$ All channels active and status byte read during communication window. Self capacitive mode, all other settings default.

[^2]:    ${ }^{1}$ Changing CS_SIZE if ATI_OFF $=0$ will change CS

[^3]:    ${ }^{2}$ ATI function will use user selected CS_SIZE and PROJ_BIAS (if applicable) and will only adjust the MULTIPLIER bits to reach the base values.

[^4]:    * Current consumption for streaming mode will differ with number of bytes read, speed and pull up resistor values

