## 100 Watt AC-DC Converters with PFC

## Input voltage range from 85... 264 V AC 1 or 2 isolated outputs up to 48 V DC 4 kV AC I/O electric strength test voltage

- Universal input voltage range
- Power factor >0.95, harmonics <IEC/EN 61000-3-2
- Input over- and undervoltage lock-out
- Efficient input filter and built-in surge and transient suppression circuitry
- Fully isolated outputs
- Outputs overload, open- and short-circuit proof
- No derating over entire operating temperature range

Safety according to IEC/EN 60950


## Summary

The S 4000/S 5000 series of AC-DC converters represents a flexible range of power supplies for use in advanced electronic systems. Features include full power factor correction, high efficiency, high reliability, low output voltage noise and excellent dynamic response to load/line changes.
The converter inputs are protected against surges and transients occurring at the source lines. An input over- and undervoltage lock-out circuitry disables the outputs if the input voltage is outside the specified range. Inrush current limitation is included preventing circuit breakers and fuses from being damaged at switch-on.
All outputs are overload, open- and short-circuit proof and are protected against overvoltages by means of a built-in suppressor diode. The outputs can be inhibited by a logic signal applied to the connector pin 18 (i). If the inhibit function is not used pin 18 must be connected to pin 14 to enable the outputs.
LED indicators display the status of the converter and allow visual monitoring of the system at any time.
Full input to output, input to case, output to case and output to output isolation is provided. The modules are designed and built according to the international safety standards IEC/EN 60950 and have been approved by the safety
agencies LGA (Germany) and UL (USA). The UL Mark for Canada has been officially recognized by regulatory authorities in provinces across Canada.
The case design allows operation at nominal load up to $71^{\circ} \mathrm{C}$ in a free air ambient temperature. If forced cooling is provided, the ambient temperature may exceed $71^{\circ} \mathrm{C}$ but the case temperature must remain below $95^{\circ} \mathrm{C}$ under all conditions.
A temperature sensor generates an inhibit signal which disables the outputs if the case temperature $T_{C}$ exceeds the limit. The outputs are automatically re-enabled when the temperature drops below the limit.
Various options are available to adapt the converters to individual applications.
The modules may either be plugged into 19" rack systems according to DIN 41494, or be chassis mounted.

## Important:

These products are intended to replace the LS 1000 and LS 2000 in order to comply with IEC/EN 61000-3-2.

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## Type Survey and Key Data

Non standard input/output configuration or special custom adaptions are available on request. See also: Commercial Information: Inquiry Form for Customized Power Supply.

Table 1: Type survey LS

| Output 1 |  | Output 2 |  | Input Voltage Range <br> $U_{i \min \ldots} U_{i \max }$ <br> 85... 255 V AC ${ }^{6}$ | Efficiency ${ }^{1}$ <br> $\eta_{\text {min }}$ <br> [\%] | Options |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $U_{0}$ nom <br> [V DC] | 10 nom <br> $[\mathrm{A}]^{2}$ | $U_{0}$ nom [V DC] | $I_{0}$ nom $[A]^{2}$ |  |  |  |
| 5.1 | 16.0 |  |  | LS 4001-7R | 77 | -9 |
| 12.0 | 8.0 |  |  | LS 4301-7R | 81 | E |
| 15.0 | 6.5 |  |  | LS 4501-7R | 83 | D |
| 24.0 | 4.2 |  |  | LS 4601-7R | 83 | V 5 |
| $24.0{ }^{3}$ | 4.0 |  |  | LS 5320-7R | 81 | P |
| $30.0{ }^{3}$ | 3.2 |  |  | LS 5540-7R | 81 | T1 |
| $48.0^{3}$ | 2.0 |  |  | LS 5660-7R | 81 | $\begin{aligned} & \text { B1 } \\ & \text { B2 } \end{aligned}$ |
| 12.0 | 4.0 | $12.0{ }^{4}$ | 4.0 | LS 5320-7R | 81 |  |
| 15.0 | 3.2 | $15.0{ }^{4}$ | 3.2 | LS 5540-7R | 81 |  |
| 24.0 | 2.0 | $24.0{ }^{4}$ | 2.0 | LS 5660-7R | 81 |  |

${ }^{1}$ Efficiency at $U_{i \text { nom }}$ and $I_{0}$ nom.
${ }^{2}$ If the output voltages are increased above $U_{0}$ nom via R-input control, option P setting, remote sensing or option T , the output currents must be reduced accordingly so that $P_{\mathrm{o}}$ nom is not exceeded.
${ }^{3}$ Series connection of output 1 and 2 , see: $R$-Function for different output configurations.
${ }^{4}$ Second output semi-regulated.
5 Option V for LS 4000 types with 5 V outputs.
${ }^{6}$ For DC-input please ask your local Power-One partner.

## Type Key

## Type Key

LS 5540-7ERPDVTB1
Input voltage range $U_{i}$ : 85 ... 255 VAC L
Series . S
Number of outputs ( 4 for single, 5 for double outputs) 4... 5
Single output units:
Nominal voltage output 1 (main output), $U_{01}$ nom
$\qquad$
$\qquad$
$\qquad$
15 V
4, 5
24 V ............................................................ 6
other voltages .......................................... 7, 8
Other specifications for single output modules ....... 01... 99
Symmetrical double output units:
Nominal voltage output 1/output 2, $U_{01 / 2 \text { nom }}$
$12 \mathrm{~V} / 12 \mathrm{~V}^{1}$ (24 V series conn.) .................. 20
$15 \mathrm{~V} / 15 \mathrm{~V}^{1}$ (30 V series conn.) ................... 40
$24 \mathrm{~V} / 24 \mathrm{~V}^{1}$ (48 V series conn.) .................. 60
other symmetrical voltages ................. 70... 99
Operational ambient temperature range $T_{\mathrm{A}}$ :

```
\(-25 . . .71^{\circ} \mathrm{C}\) \(-7\)
```

$\qquad$
customer specific
0...-6

Auxiliary functions and options:
Inrush current limitation ...............................E
Output voltage control input ........................ R ${ }^{2}$
Potentiometer (output voltage adjustment) .. $\mathrm{P}^{2}$
Save data signal (DO...DD, to be specified) $D^{3}$
ACFAIL signal (V2, V3) ................................V ${ }^{3,4}$
Current sharing ............................................ T
Cooling plate standard case ......................B1
Cooling plate for longe case $220 \mathrm{~mm} . . . . . .$. B2
${ }^{1}$ External wiring of main and second output depending upon the desired output configuration (see: $R$-Function for different output configurations).
${ }^{2}$ Feature R excludes option P and vice versa.
${ }^{3}$ Option D excludes option $V$ and vice versa.
${ }^{4}$ Option V available for LS 4000 types with 5 V output.

## Functional Description

The input voltage is fed via an input fuse, an input filter, a rectifier and an inrush current limiter to a single transistor boost converter. This converter provides a sinusoidal input current (IEC/EN 61000-3-2, class D equipment) and sources a capacitor with a voltage of $360-370 \mathrm{~V}$ DC. This capacitor sources a single transistor forward converter.
Each output is powered by a separate secondary winding of the main transformer. The resultant voltages are rectified and their ripples smoothed by a power choke and an output
filter. The control logic senses the main output voltage $U_{01}$ and generates, with respect to the maximum admissible output currents, the control signal for the primary switching transistor.
The second output of double output units is controlled by the main output, but has independent current limiting. If the main output is driven into current limitation, the second output voltage will fall as well and vice versa.

Fig. 1


Block diagram of single output converters LS 4000

Fig. 2


Block diagram of symmetrical double output converters LS 5000
${ }^{1}$ Transient suppressor (VDR)
${ }^{2}$ Inrush current limiter (NTC or Opt. E), -9 versions exclude the NTC
${ }^{3}$ Input fuse
${ }^{4}$ Hold-up capacitor

## Electrical Input Data

General Conditions

- $T_{\mathrm{A}}=25^{\circ} \mathrm{C}$, unless $T_{\mathrm{C}}$ is specified.
- Pin 18 connected to pin $14, U_{0}$ adjusted to $U_{0}$ nom (option P); R input not connected.
- Sense line pins S+ and S- connected to Vo+ and Vo- respectively.

Table 2: Input data

| Input |  |  | LS |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Characteristics |  | Conditions | min | typ | max |  |
| $U_{i}$ | Operating Input voltage | $\begin{aligned} & I_{\mathrm{o}}=0 \ldots I_{\mathrm{onom}} \\ & T_{\mathrm{C} \text { min }} \ldots T_{\mathrm{C} \text { max }} \end{aligned}$ | 85 |  | 255 | V AC ${ }^{3,4}$ |
| $U_{\text {i nom }}$ | Nominal Input voltage |  | 230 |  |  |  |
| $l_{i}$ | Input current | $U_{\text {i nom }}, I_{\text {o nom }}{ }^{1}$ | 0.55 |  |  | $\mathrm{A}_{\text {rms }}$ |
| $P_{\text {i } 0}$ | No-load input power | $U_{\mathrm{i} \text { min } \ldots} \ldots U_{\mathrm{imax}}$ |  | 9.0 | 10 | W |
| $P_{\mathrm{i} \text { inh }}$ | Idle input power | unit inhibited |  | 3.5 | 5 |  |
| $R_{\text {i }}$ | Input resistance | $T_{\mathrm{C}}=25^{\circ} \mathrm{C}$ | 480 |  |  | $\mathrm{m} \Omega$ |
| $R_{\text {NTC }}$ | NTC resistance ${ }^{2}$ |  | 3200 | 4000 |  |  |
| $C_{i}$ | Input capacitance |  | 80 | 100 | 120 | $\mu \mathrm{F}$ |
| $U_{i \text { RFI }}$ | Conducted input RFI | EN 55022 <br> $U_{\text {i nom }}, I_{\mathrm{o}}$ nom | B |  |  |  |
|  | Radiated input RFI |  | B |  |  |  |
| $U_{i}$ abs | Input voltage limits without damage |  | -400 |  | 400 | V DC |
|  |  |  | -400 |  | 400 | $\mathrm{V}_{\mathrm{p}}$ |

${ }^{1}$ With double output modules, both outputs loaded with $I_{0}$ nom.
${ }^{2}$ Valid for -7 versions with NTC, (-9 versions exclude the NTC). Initial switch-on cycle. Subsequent switch-on/off cycles increase the inrush current peak value.
${ }^{3} \mathrm{AC}$ frequency range $47 . .63 \mathrm{~Hz}$.
${ }^{4}$ For DC-input please ask your local Power-One partner.

## Input Fuse

A fuse mounted inside the converter protects the module against severe defects. (If operated from a DC-source this fuse may not fully protect the module when the input voltage exceeds 200 V DC! In applications where the converters operate at source voltages above 200 V DC an external fuse or a circuit breaker at system level should be installed!)

Table 3: Fuse Specification

| Module | Fuse type | Fuse rating |  |
| :--- | :--- | :--- | :--- |
| LS $^{1}$ | slow-blow | SP T | $4 \mathrm{~A}, 250 \mathrm{~V}$ |

${ }^{1}$ Fuse size $5 \times 20 \mathrm{~mm}$

## Input Under-/Overvoltage Lock-out

If the input voltage remains below approx. 65 V AC or exceeds approx. 280 V AC an internally generated inhibit signal disables the output(s). When checking this function the absolute maximum input voltage rating $U_{i \text { abs }}$ should be considered! Between $U_{i \min }$ and the undervoltage lock-out level the output voltage may be below the value defined in table: Output data (see: Technical Information: Measuring and Testing).

## Input Transient Protection

A VDR together with the input fuse and a symmetrical input filter form an effective protection against high input transient voltages.

## Reverse Polarity Protection

Should the input voltage to the unit be supplied from a DC source the built-in bridge rectifier provides reverse polarity protection. (For DC-input operation, please consult your local Power-One partner.)

## Inrush Current Limitation

The modules of the versions -7, incorporate an NTC resistor in the input circuitry which - at initial turn on - reduces the peak inrush current value by a factor of $5 \ldots 10$ to protect connectors and switching devices from damage. Subsequent switch-on cycles within short periods will cause an increase of the peak inrush current value due to the warm-ing-up of the NTC resistor. See also: E option.

## Inrush Current Peak Value

The inrush current peak value (initial switch-on cycle) can be determined by following calculation:

$$
l_{\mathrm{inr} \mathrm{p}}=\frac{U_{\mathrm{irms}} \cdot \sqrt{2}}{\left(R_{\mathrm{s} \text { ext }}+R_{\mathrm{i}}+R_{\mathrm{NTC}}\right)}
$$



Fig. 3
Equivalent circuit diagram for input impedance.

Input Inrush Current Characteristic


Fig. 4
Theoretical input inrush current versus time at $U_{\mathrm{i}} 255 \mathrm{~V}_{\mathrm{rms}}$ and $115 \mathrm{~V}_{\mathrm{rms}}, R_{\mathrm{ext}}=0$.

## Harmonic Currents

The harmonic distortion is well below the limits specified in IEC/EN 61000-3-2, class D.


Fig. 5
Harmonic currents at the input, IEC/EN 61000-3-2, class $D$. $U_{\mathrm{i}}=U_{\mathrm{i} \text { nom }}, I_{0}=I_{\mathrm{o}}$ nom.

## Static Input current Characteristic



Fig. 6
Input current versus input voltage at $I_{0}$ nom

## Power Factor, Harmonics

Power factor correction is achieved by controlling the input current waveform synchronously with the input voltage waveform. The power factor control is active under all operating conditions.


Fig. 7
Power factor versus output current at $U_{\mathrm{i}} 230$ V AC and 85 V AC.

## Electrical Output Data

General Conditions
$-T_{\mathrm{A}}=25^{\circ} \mathrm{C}$, unless $T_{\mathrm{C}}$ is specified.

- Pin 18 (i) connected to pin 14 ( $\mathrm{S}-/ \mathrm{Vo1-}$ ), $U_{0}$ adjusted to $U_{0}$ nom (option P), R input not connected.
- Sense line pins 12 (S+) and 14 (S-) connected to pins 4 (Vo1+) and 8 (Vo1-) respectively.

Table 4a: Output data single output modules

${ }^{1}$ If the output voltages are increased above $U_{0}$ nom through R-input control, option P setting, remote sensing or option T , the output currents should be reduced accordingly so that $P_{\text {o nom }}$ is not exceeded.
2 See: Output Voltage Regulation of Single or Double Output Modules with Outputs 1 and 2 Connected in Series.
${ }^{3}$ See: Typical dynamic load regulation of $U_{01}$ and $U_{02}$
${ }^{4}$ Negative temperature coefficient ( $0 . . .-3 \mathrm{mV} / \mathrm{cell}$ and K ) available on request.
${ }^{5}$ Measured according to IEC/EN 61204 sub clause 3.10 with a probe acc. to annex A of the same standards. (See:Technical Information: Measuring and Testing)

Table 4b: Output data double output modules

| Output <br> (Outputs connected in Series) |  |  |  | $\begin{gathered} \text { LS } 5320 \\ 24 \mathrm{~V}(2 \times 12 \mathrm{~V}) \end{gathered}$ |  |  | $\begin{gathered} \text { LS } 5540 \\ 30 \mathrm{~V}(2 \times 15 \mathrm{~V}) \end{gathered}$ |  |  | $\begin{gathered} \text { LS } 5660 \\ 48 \mathrm{~V}(2 \times 24 \mathrm{~V}) \end{gathered}$ |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Characteristics |  |  | Conditions | min | typ | max | min | typ | max | min | typ | max |  |
| $U_{0}$ | Output voltage ${ }^{2}$ |  | $U_{\text {i nom, }} I_{\text {o nom }}$ |  | 24.0 |  |  | 30.0 |  |  | 48.0 |  | V |
| $U_{\text {op }}$ | Overvoltage protection (supressor diode) |  |  | 38 |  |  | 48 |  |  | 74 |  |  |  |
| $I_{\text {o nom }}$ | Output current ${ }^{1}$ |  | $\begin{aligned} & U_{\mathrm{i} \text { min } \ldots} . U_{\mathrm{imax}} \\ & T_{\mathrm{C} \text { min } \ldots} T_{\mathrm{C} \text { max }} \end{aligned}$ | 4.0 |  |  | 3.2 |  |  | 2.0 |  |  | A |
| 10 L | Output current limit ${ }^{4}$ |  | $U_{i \min \ldots} \ldots U_{\mathrm{i} \text { max }}$ | 4.2 |  |  | 3.4 |  |  | 2.1 |  |  |  |
| $u_{0}{ }^{7}$ | Output voltage noise ${ }^{3}$ | Low frequency | $U_{\text {i nom }}, I_{\text {o nom }}$ IEC/EN 61204 $B W=20 \mathrm{MHz}$ | 3 |  |  | 3 |  |  | 5 |  |  | $\mathrm{mV}_{\mathrm{pp}}$ |
|  |  | Switching freq. |  | 15 |  |  | 15 |  |  | 20 |  |  |  |
|  |  | Total |  | 100 |  |  | 100 |  |  | 150 |  |  |  |
| $\Delta U_{0} \mathrm{U}$ | Static line regulation |  | $\begin{aligned} & U_{i \min \ldots} U_{\mathrm{i} \max } \\ & I_{0 \text { nom }} \end{aligned}$ | $\pm 12$ |  |  | $\pm 15$ |  |  | $\pm 24$ |  |  | mV |
| $\Delta U_{0}$ I | Static load regulation |  | $\begin{aligned} & U_{i \text { nom }} I_{0}= \\ & (0.1 . .1) I_{\text {onom }} \end{aligned}$ | 40 |  |  | 60 |  |  | 96 |  |  |  |
| $u_{0}{ }^{5}$ | Dynamic load regulat. | Voltage deviation | $U_{i \text { nom }}, I_{0}=$ <br> $I_{0 \text { nom }} \leftrightarrow 1 / 2 I_{0 \text { nom }}$ <br> IEC/EN 61204 | $\pm 250$ |  |  | $\pm 200$ |  |  | $\pm 150$ |  |  |  |
| $t_{\mathrm{d}}{ }^{5}$ |  | Recovery time |  | 0.3 |  |  | 0.3 |  |  | 0.3 |  |  | ms |
| $\alpha_{\text {Uo }}$ | Temperature coefficient of output voltage ${ }^{6}$ |  | $\begin{aligned} & U_{i \min \ldots} U_{i \max } \\ & 0 \ldots I_{\text {onom }} \end{aligned}$ | -2.2 |  |  | -2.2 |  |  | -2.6 |  |  | $\mathrm{mV} / \mathrm{K}$ |

${ }^{1}$ If the output voltages are increased above $U_{o n o m}$ through R-input control, option P setting, remote sensing or option T , the output currents should be reduced accordingly so that $P_{\mathrm{o} \text { nom }}$ is not exceeded.
${ }^{2}$ Series connection for $U_{0}$ nom $=24 \mathrm{~V}, 30 \mathrm{~V}$ or 48 V , see: $R$-Function for different output configurations.
${ }^{3}$ Shortest possible wiring for series connection at the connector.
${ }^{4}$ See: Output Voltage Regulation of Single or Double Output Modules with Outputs 1 and 2 Connected in Series.
${ }^{5}$ See: Typical dynamic load regulation of $U_{01}$ and $U_{02}$.
${ }^{6}$ Negative temperature coefficient ( $0 \ldots-3 \mathrm{mV} /$ cell and K ) available on request.
${ }^{7}$ Measured according to IEC/EN 61204 sub clause 3.10 with a probe acc. to annex A of the same standards. (See: Technical Information: Measuring and Testing)

Table 4c: Output data double output modules

| Output <br> (Outputs independently loaded) ${ }^{1}$ |  |  |  | $\begin{gathered} \text { LS } 5320 \\ 12 \mathrm{~V} / 12 \mathrm{~V} \end{gathered}$ |  |  |  |  | $\begin{gathered} \text { LS } 5540 \\ 15 \mathrm{~V} / 15 \mathrm{~V} \end{gathered}$ |  |  |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Characteristics |  |  | Conditions | Output 1 |  | Output 2 |  |  | Output 1 |  | Output 2 |  |  |  |
|  |  |  | min typ | typ max | min | typ | max | min | typ max | min | typ | max |  |
| $U_{0}$ | Output voltage |  |  | $U_{\text {i nom }}, I_{0}$ nom ${ }^{2}$ | 11.93 | 12.07 | 11.82 |  | 12.18 | 14.91 | 15.09 | 14.78 |  | 15.23 | V |
| $U_{\text {op }}$ | Overvoltage protection (supressor diode) |  |  | 19 |  | 19 |  |  | 24 |  | 24 |  |  |  |  |
| $I_{\text {o nom }}$ | Output current ${ }^{3}$ |  | $\begin{aligned} & U_{\mathrm{i} \min \ldots} . U_{\mathrm{imax}} \\ & T_{\mathrm{C} \text { min } \ldots} T_{\mathrm{C} \text { max }} \end{aligned}$ | 4.0 |  | 4.0 |  |  | 3.2 |  | 3.2 |  |  | A |  |
| $I_{0 L}$ | Output current limit ${ }^{4}$ |  | $U_{i \text { min } \ldots} U_{\mathrm{i} \text { max }}$ | 4.2 |  | 4.2 |  |  | 3.4 |  | 3.4 |  |  |  |  |
| $u_{0}{ }^{8}$ | Output voltage noise | Low frequency | $\begin{aligned} & U_{\text {inom, }} I_{0 \text { nom }} \\ & \text { IEC/EN } 61204 \\ & \text { BW }=20 \mathrm{MHz} \end{aligned}$ | 3 |  | 3 |  |  | 3 |  | 3 |  |  | $m V_{p p}$ |  |
|  |  | Switching freq. |  | 12 |  | 12 |  |  | 10 |  | 10 |  |  |  |  |
|  |  | Total |  | 80 |  | 40 |  |  | 100 |  | 40 |  |  |  |  |
| $\Delta U_{0} \mathrm{u}$ | Static line regulation |  | $\begin{aligned} & U_{i \text { min } \ldots} U_{i \text { nom }} \\ & U_{i \text { nom } \ldots} U_{i \operatorname{inax}} \\ & I_{\text {onom }} \end{aligned}$ | $\pm 12$ |  | $\pm 12$ |  |  | $\pm 15$ |  | $\pm 15$ |  |  | mV |  |
| $\Delta U_{0}$ I | Static load regulation |  | $\begin{aligned} & U_{\text {inom }}, I_{0}= \\ & (0.1 \ldots 1) I_{0 \text { nom }}{ }^{5} \end{aligned}$ | 48 |  | 5 |  |  | 60 |  | 5 |  |  |  |  |
| $u_{0} d^{6}$ | Dynamic load regulat. | Voltage deviation | $\begin{aligned} & U_{\text {i nom }} I_{\circ}= \\ & I_{\text {onom }} \leftrightarrow 1 / 2 I_{0 \text { nom }} \\ & \text { IEC/EN } 61204 \end{aligned}$ | $\pm 100$ |  |  |  |  | $\pm 100$ |  |  |  |  |  |  |
| $t_{\mathrm{d}}{ }^{6}$ |  | Recovery time |  | 0.3 |  |  |  |  | 0.3 |  |  |  |  | ms |  |
| $\alpha_{\text {Uo }}$ | Temperature coefficient of output voltage ${ }^{7}$ |  | $\begin{aligned} & U_{i \min \ldots} \ldots U_{i \max } \\ & 0 \ldots I_{\text {onom }} \end{aligned}$ | -1.5 |  |  |  |  | -1.5 |  |  |  |  | $\mathrm{mV} / \mathrm{K}$ |  |

Table 4d: Output data double output modules

| Output <br> (Outputs independently loaded) ${ }^{1}$ |  |  |  | $\begin{aligned} & \text { LS } 5660 \\ & 24 \text { V/24 V } \end{aligned}$ |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Characteristics |  |  | Conditions | Output 1 | Output 2 |  |  |
|  |  |  | min typ max | min | typ max |  |
| $U_{0}$ | Output voltage |  |  | $U_{\text {i nom, }}, I_{0 \text { nom }}{ }^{2}$ | $23.86 \quad 24.14$ | 23.64 | 24.36 | V |
| $U_{\text {op }}$ | Overvoltage protection (supressor diode) |  |  | 37 | 37 |  |  |  |
| $I_{0}$ nom | Output current ${ }^{3}$ |  | $\begin{aligned} & U_{\mathrm{i} \text { min }} \ldots U_{\mathrm{imax}} \\ & T_{\mathrm{C} \text { min }} \ldots T_{\mathrm{C} \text { max }} \end{aligned}$ | 2.0 | 2.0 |  | A |  |
| $I_{0}$ L | Output current limit ${ }^{4}$ |  | $U_{\mathrm{i} \text { min } \ldots} . U_{\mathrm{i} \text { max }}$ | 2.1 | 2.1 |  |  |  |
| $u_{0}{ }^{8}$ | Output voltage noise | Low frequency | $U_{i}$ nom,$I_{\text {onom }}$ IEC/EN 61204$\mathrm{BW}=20 \mathrm{MHz}$ | 3 | 3 |  | $m V_{p p}$ |  |
|  |  | Switching freq. |  | 10 | 10 |  |  |  |
|  |  | Total |  | 100 | 40 |  |  |  |
| $\Delta U_{0} \mathrm{U}$ | Static line regulation |  | $\begin{aligned} & U_{i \min \ldots} \ldots U_{i \text { nom }}, \\ & U_{\text {i nom }} . . U_{\mathrm{i} \max } \\ & I_{\text {onom }} \end{aligned}$ | $\pm 24$ | $\pm 24$ |  | mV |  |
| $\Delta U_{0}$ I | Static load regulation |  | $\begin{aligned} & U_{i \text { nom, }}, I_{0}= \\ & (0.1 \ldots 1) I_{0 \text { nom }}{ }^{5} \end{aligned}$ | 96 | 5 |  |  |  |
| $u_{o d}{ }^{6}$ | Dynamic load regulat. | Voltage deviation | $\begin{aligned} & U_{\text {i nom }}, I_{0}= \\ & I_{\text {onom }} \leftrightarrow 1 / 2 I_{\text {o nom }} \\ & \text { IEC/EN } 61204 \end{aligned}$ | $\pm 80$ |  |  |  |  |
| $t_{\mathrm{d}}{ }^{6}$ |  | Recovery time |  | 0.3 |  |  | ms |  |
| $\alpha_{\text {Uo }}$ | Temperature coefficient of output voltage ${ }^{7}$ |  | $\begin{aligned} & U_{i \min \ldots} U_{i \max } \\ & 0 \ldots I_{\mathrm{onom}} \end{aligned}$ | -0.5 |  |  | $\mathrm{mV} / \mathrm{K}$ |  |

${ }^{1}$ Depending upon the desired output configuration the wiring should be made as shown in: R-Function for different output configurations.
${ }^{2}$ Same conditions for both outputs.
${ }^{3}$ If the control voltages are increased above $U_{\text {o nom }}$ via R-input control, option Psetting, remote sensing or option T, the output currentsshould be reduced accordingly so that $P_{\text {o nom }}$ is not exceeded.
${ }^{4}$ See: Output Voltage Regulation of Single or Double Output Modules with Outputs 1 and 2 Connected in Series.
${ }^{5}$ Condition for specified output. Other output loaded with constant current $I_{0}=I_{0}$ nom. See: Output voltage regulation of double output units.
${ }^{6}$ See: Typical dynamic load regulation of $U_{01}$ and $U_{02}$.
${ }^{7}$ Negative temperature coefficient ( $0 . \ldots .-3 \mathrm{mV} / \mathrm{cell}$ and K ) available on request.
${ }^{8}$ Measured according to IEC/EN 61204 sub clause 3.10 with a probe acc. to annex A of the same standards.

## Thermal Considerations

If a converter is located in free, quasi-stationary air (convection cooling) at the indicated maximum ambient temperature $T_{\text {Amax }}$ (see table: Temperature specifications) and is operated at its nominal input voltage and output power, the temperature measured at the Measuring point of case temperature $T_{\mathrm{C}}$ (see: Mechanical Data) will approach the indicated value $T_{\mathrm{C} \text { max }}$ after the warm-up phase. However, the relationship between $T_{\mathrm{A}}$ and $T_{\mathrm{C}}$ depends heavily on the conditions of operation and integration into a system. The thermal conditions are influenced by input voltage, output current, airflow and temperature of surrounding components and surfaces. $T_{\mathrm{A} \text { max }}$ is therefore, contrary to $T_{\mathrm{C} \text { max }}$, an indicative value only.
Caution: The installer must ensure that under all operating conditions $T_{C}$ remains within the limits stated in the table: Temperature specifications.

Notes: Sufficient forced cooling or an additional heat sink allows $T_{\mathrm{A}}$ to be higher than $71^{\circ} \mathrm{C}\left(\mathrm{e} . \mathrm{g} .85^{\circ} \mathrm{C}\right)$ if $T_{\mathrm{C} \text { max }}$ is not exceeded.
For -7 or -9 units at an ambient temperature $T_{\mathrm{A}}$ of $85^{\circ} \mathrm{C}$ with only convection cooling, the maximum permissible current for each output is approx. $40 \%$ of its nominal value as per figure.


Fig. 8
Output current derating versus temperature for -7 and -9 units.

## Thermal Protection

A temperature sensor generates an internal inhibit signal which disables the outputs if the case temperature exceeds $T_{\mathrm{C} \text { max }}$. The outputs are automatically re-enabled if the temperature drops below this limit.
It is recommended that continuous operation under simultaneous extreme worst case conditions of the following three parameters be avoided: Minimum input voltage, maximum output power and maximum temperature.

## Output Protection

Each output is protected against overvoltage which could occur due to a failure of the control circuit by means of a voltage suppressor diode which, under worst case conditions, may become a short circuit. The suppressor diodes are not designed to withstand externally applied overvoltages. Overload at any of the two outputs will cause a shut-down of both outputs. A red LED indicates the overload condition.

## Parallel or Series Connection of Units

Single or double output units with equal nominal output voltage can be connected in parallel without any precautions using option T .
With option T (current sharing), all units share the current approximately equally.
Single output units and/or main and second outputs of double output units can be connected in series with any other (similar) output.

## Note:

- Parallel connection of double output units should always include both, main and second output to maintain good regulation of both outputs.
- Not more than 5 units should be connected in parallel.
- Series connection of second outputs without involving their main outputs should be avoided as regulation may be poor.
- The maximum output current is limited by the output with the lowest current limitation if several outputs are connected in series.


## Output Voltage Regulation of Single or Double Output

 Modules with Outputs 1 and 2 Connected in Series

Fig. 9
$U_{01}$ vs. $I_{01}$ (typ.) of single output units

## Output Voltage Regulation of Double Output Modules

Output 1 is under normal conditions regulated to $U_{01}$ nom, independent of the output currents.
$U_{02}$ is dependent upon the load distribution. If both outputs are loaded with more than $10 \%$ of $I_{0}$ nom, the deviation of $U_{02}$ remains within $\pm 5 \%$ of the value of $U_{01}$. The following 3 figures show the regulation with varying load distribution. If $I_{01}=I_{02}$ or the two outputs are connected in series, the deviation of $U_{02}$ remains within $\pm 1 \%$ of the value of $U_{01}$ provided that a total load of more than $10 \%$ of $I_{o n o m}$ is applied.
Two outputs of a single S 5000 module connected in parallel will behave like the output of a $S 4000$ module; the paralleled output is fully regulated. No precautions are necessary in using the R-input and the test sockets.


Fig. 10
LS 5320: $\Delta U_{02}$ (typ.) vs. $l_{02}$ with different $l_{01}$.


Fig. 11
LS 5540: $\Delta U_{02}$ (typ.) vs. $l_{02}$ with different $l_{01}$.
[V] $U_{02}$


Fig. 12
LS 5660: $\Delta U_{02}$ (typ.) vs. $I_{02}$ with different $I_{01}$.
Switching Frequency versus Load


Fig. 13
Switching frequency versus load. (The boost converter at the input stage has a fixed frequency of 100 kHz )

## Dynamic Load Regulation



Fig. 14
Typical dynamic load regulation of $U_{01}$ and $U_{02}$.

## Hold-up Time versus Output Power



Fig. 15
Hold-up time $t_{\mathrm{h}}$ versus output power.

## Efficiency versus Load



Fig. 16
Efficiency versus load at $U_{i} ; 230$ V AC and 85 V AC

## Auxiliary Functions

## i Inhibit for Remote On and Off

Note: With open i input: Output is disabled ( $U_{0}=0$ off).
The outputs of the module may be enabled or disabled by means of a logic signal (TTL, CMOS, etc.) applied between the inhibit input $i$ and the negative pin of output 1 (Vo1-). In systems with several units, this feature can be used, for example, to control the activation sequence of the converters. If the inhibit function is not required, connect the inhibit pin 18 to pin 14 to enable the outputs (active low logic, fail safe). For output response refer to: Hold-up Time and Output Response.


Fig. 17
Definition of $U_{\text {inh }}$ and $l_{\text {inh }}$.


Fig. 18
Typical inhibit current $l_{\text {inh }}$ versus inhibit voltage $U_{\text {inh }}$
Table 5: Inhibit characteristics



Fig. 19
Output response as a function of inhibit control

Programmable Output Voltage (R-Function)
As a standard feature, the modules offer an adjustable output voltage, identified by letter R in the type designation. The control input $R$ (pin 16) accepts either a control voltage $U_{\text {ext }}$ or a resistor $R_{\text {ext }}$ to adjust the desired output voltage. When not connected, the control input automatically sets the output voltage to $U_{0}$ nom.
a) Adjustment by means of an external control voltage $U_{\text {ext }}$ between pin $16(\mathrm{R})$ and pin 14:
The control voltage range is $0 . . .2 .75 \mathrm{~V}$ DC and allows an output voltage adjustment in the range of approximately $0 . . .110 \% U_{0}$ nom.
$U_{\text {ext }}=\frac{U_{0}}{U_{0 \text { nom }}} \cdot 2.5 \mathrm{~V}$ (approximate formula)
b) Adjustment by means of an external resistor:

Depending upon the value of the required output voltage the resistor shall be connected
either: Between pin 16 and pin $14\left(U_{0}<U_{0}\right.$ nom $)$ to achieve an output voltage adjustment range of approximately $0 . . .100 \% U_{0}$ nom
or: Between pin 16 and pin $12\left(U_{0}>U_{0}\right.$ nom $)$ to achieve an output voltage adjustment range of approximately 100...110\% Uo nom.

## Warning:

- Uext shall never exceed 2.75 V DC.
- The value of $R_{\text {ext }}^{\prime}$ shall never be less than the lowest value as indicated in table $R_{\text {ext }}^{\prime}$ (for $U_{0}>U_{0 \text { nom }}$ ) to avoid damage to the unit!


## Remarks:

- The R-Function excludes option P (output voltage adjustment by potentiometer).
- If the output voltages are increased above $U_{0 \text { nom }}$ via Rinput control, option $P$ setting, remote sensing or option T , the output current(s) should be reduced accordingly so that $P_{\mathrm{o}}$ nom is not exceeded.
- The R-input (as well as option $P$ ) is related to the main output.
- With double output units the second output follows the value of the controlled main output. Resistor values as indicated for the single output units should be used.
- For correct output voltage adjustment of double output units the external wiring of the outputs should be according to fig.: R-Function for different output configuration depending upon the desired output configuration.
- In case of parallel connection the output voltages should be individually set within a tolerance of $1 . . .2 \%$.


Fig. 20
Output voltage control for single output units LS 4000 by means of the $R$ input

Table 7a: $R_{\text {ext }}$ for $U_{0}<U_{0}$ nom; approximative values ( $U_{\mathrm{i} \text { nom }}, I_{0}$ nom, series $E 96$ resistors); $R_{\text {ext }}^{\prime}=\infty$

| $U_{\text {o nom }}=5.1 \mathrm{~V}$ |  | $U_{\text {o nom }}=12 \mathrm{~V}$ |  |  | $U_{\text {o nom }}=15 \mathrm{~V}$ |  |  | $U_{\text {o nom }}=\mathbf{2 4} \mathbf{V}$ |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $U_{0}(\mathrm{~V})$ | $R_{\text {ext }}[\mathrm{k} \Omega]$ | $U_{0}[\mathrm{~V}]^{1}$ |  | $R_{\text {ext }}[\mathrm{k} \Omega]$ | $U_{0}[\mathrm{~V}]^{1}$ |  | $R_{\text {ext }}[\mathrm{k} \Omega$ ] | $U_{0}[\mathrm{~V}]^{1}$ |  | $R_{\text {ext }}[\mathrm{k} \Omega]$ |
| 0.5 | 0.432 | 2 | 4 | 0.806 | 2 | 4 | 0.619 | 4 | 8 | 0.806 |
| 1.0 | 0.976 | 3 | 6 | 1.33 | 4 | 8 | 1.47 | 6 | 12 | 1.33 |
| 1.5 | 1.65 | 4 | 8 | 2 | 6 | 12 | 2.67 | 8 | 16 | 2 |
| 2.0 | 2.61 | 5 | 10 | 2.87 | 8 | 16 | 4.53 | 10 | 20 | 2.87 |
| 2.5 | 3.83 | 6 | 12 | 4.02 | 9 | 18 | 6.04 | 12 | 24 | 4.02 |
| 3.0 | 5.76 | 7 | 14 | 5.62 | 10 | 20 | 8.06 | 14 | 28 | 5.62 |
| 3.5 | 8.66 | 8 | 16 | 8.06 | 11 | 22 | 11 | 16 | 32 | 8.06 |
| 4.0 | 14.7 | 9 | 18 | 12.1 | 12 | 24 | 16.2 | 18 | 36 | 12.1 |
| 4.5 | 30.1 | 10 | 20 | 20 | 13 | 26 | 26.1 | 20 | 40 | 20 |
| 5.0 | 200 | 11 | 22 | 42.2 | 14 | 28 | 56.2 | 22 | 44 | 44.2 |

Table 7b: $R_{\text {ext }}^{\prime}$ for $U_{0}>U_{0}$ nom; approximative values ( $U_{\mathrm{i}}$ nom,,$I_{0 \text { nom }}$, series $E 96$ resistors); $R_{\text {ext }}=\infty$

| $U_{0}$ nom $=5.1 \mathrm{~V}$ |  | $U_{\text {o nom }}=12 \mathrm{~V}$ |  |  | $U_{\text {o nom }}=15 \mathrm{~V}$ |  |  | $U_{\text {o nom }}=24 \mathrm{~V}$ |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $U_{0}[\mathrm{~V}]$ | $R_{\text {ext }}^{\prime}[\mathrm{k} \Omega]$ | $U_{0}[\mathrm{~V}]^{1}$ |  | $R_{\text {ext }}^{\prime}[\mathrm{k} \Omega]$ | $U_{0}[\mathrm{~V}]^{1}$ |  | $R_{\text {ext }}^{\prime}[\mathrm{k} \Omega]$ | $\mathrm{U}_{0}[\mathrm{~V}]^{1}$ |  | $R_{\text {ext }}^{\prime}[\mathrm{k} \Omega]$ |
| 5.15 | 432 | 12.1 | 24.2 | 1820 | 15.2 | 30.4 | 1500 | 24.25 | 48.5 | 3320 |
| 5.2 | 215 | 12.2 | 24.4 | 931 | 15.4 | 30.8 | 768 | 24.5 | 49.0 | 1690 |
| 5.25 | 147 | 12.3 | 24.6 | 619 | 15.6 | 31.2 | 523 | 24.75 | 49.5 | 1130 |
| 5.3 | 110 | 12.4 | 24.8 | 475 | 15.8 | 31.6 | 392 | 25.0 | 50.0 | 845 |
| 5.35 | 88.7 | 12.5 | 25.0 | 383 | 16.0 | 32.0 | 316 | 25.25 | 50.5 | 698 |
| 5.4 | 75 | 12.6 | 25.2 | 316 | 16.2 | 32.4 | 267 | 25.5 | 51.0 | 590 |
| 5.45 | 64.9 | 12.7 | 25.4 | 274 | 16.4 | 32.8 | 232 | 25.75 | 51.5 | 511 |
| 5.5 | 57.6 | 12.8 | 25.6 | 243 | 16.5 | 33.0 | 221 | 26.0 | 52.0 | 442 |
|  |  | 13.0 | 26.0 | 196 |  |  |  | 26.25 | 52.5 | 402 |
|  |  | 13.2 | 26.4 | 169 |  |  |  | 26.4 | 52.8 | 383 |

[^0]
## R-Function for different output configurations



Fig. 21a
LS 5000 with H15 connector. R-input for output voltage control. Wiring for output voltage 24 V or 30 V or 48 V with main and second output connected in series.


Fig. 21c
LS 5000 with H15 connector. R-input for output voltage control. Wiring of main and second output for two symmetrical output voltages $U_{01}$ and $U_{02}: \pm 12 \mathrm{~V}$ or $\pm 15 \mathrm{~V}$ or $\pm 24 \mathrm{~V}$.


Fig. 21e
LS 5000 with H15 connector. R-input for output voltage control. Wiring of main and second output for two output voltages $U_{01}$ and $U_{02}$ : $12 \mathrm{~V} / 12 \mathrm{~V}$ or $15 \mathrm{~V} / 15 \mathrm{~V}$ or $24 \mathrm{~V} / 24 \mathrm{~V}$, the outputs are galvanically isolated.


Fig. 21b
LS 5000 with H15 connector. R-input for output voltage control. Wiring for output voltage 12 V or 15 V or 24 V with main and second output connected in parallel.


Fig. 21d
LS 5000 with H15 connector. R-input for output voltage control. Wiring of main and second output for two output voltages $U_{01}$ and $U_{02}:+12 \mathrm{~V}$ and +24 V or +15 V and +30 V or +24 V and +48 V .
${ }^{1}$ A ceramic multilayer capacitor connected across the load reduces ripple and spikes.
${ }^{2}$ Shortest possible wiring for series connection at the female connector

## Remarks:

Double output units fitted with H 15 connectors have the output pins of the second output, pins $4 / 6$ and $8 / 10$, internally paralleled.
It is recommended that pins $4 / 6$ and $8 / 10$ be directly paralleled at the female connector as well to reduce the voltage drop across the connector.
Please note: $U_{02}$ varies depending upon its own load and the load on output 1.

Display Status of LEDs


## Battery Charging/Temperature Sensor

The LS are intended for lead acid battery charger applications. For an optimum battery charging and life expectancy of the battery an external temperature sensor may be connected to the R-input. The sensor is mounted as close as possible to the battery pole and adjusts the output voltage of the LS unit according to the temperature of the battery (which is related to the load of the battery and the ambient temperature).
Depending on the cell voltage and the temperature coefficient of the battery, different sensor types are available.
For more information please ask Power-One.


Fig. 17
Dependance of output voltage vs. temperature for defined temperature coefficient.

Fig. 22
LEDs "OK", "i" and " $\mathrm{l}_{0}$ " ${ }^{\text {status versus input voltage }}$ Conditions: $I_{0} \leq I_{\text {onom }}, T_{\mathrm{C}} \leq T_{\mathrm{C} \text { max }}, U_{\text {inh }} \leq 0.8 \mathrm{~V}$
$U_{\mathrm{i} \mathrm{uv}}=$ undervoltage lock-out, $U_{\mathrm{i} \text { ov }}=$ overvoltage lock-out

LEDs "OK" and " $\mathrm{O}_{\mathrm{L}}$ "status versus output current Conditions: $U_{\mathrm{i} \text { min } \ldots} U_{\mathrm{i} \text { max }}, T_{\mathrm{C}} \leq T_{\mathrm{C} \text { max }}, U_{\text {inh }} \leq 0.8 \mathrm{~V}$

LED "i"versus case temperature
Conditions: $U_{i \min } \ldots U_{\mathrm{i} \text { max }}, I_{0} \leq I_{\text {nom }}, U_{\text {inh }} \leq 0.8 \mathrm{~V}$

LED "i"versus Uinh
Conditions: $U_{\mathrm{i} \text { min } \ldots} U_{\mathrm{i} \text { max }}, I_{\mathrm{O}} \leq I_{\mathrm{O} \text { nom }}, T_{\mathrm{C}} \leq T_{\mathrm{C} \text { max }}$

Test Sockets (Main output only )
Test sockets for measuring the output voltage $U_{01}$ are located at the front of the module. The positive test socket is protected by a series resistor (see: Functional Description, block diagrams). The voltage measured at the test sockets is approximately 30 mV lower than the value measured at the output terminals.
In case of double output units externally connected in series for $U_{0}=24 \mathrm{~V}, 30 \mathrm{~V}$ or 48 V the monitored output voltage is $12 \mathrm{~V}, 15 \mathrm{~V}$ or 24 V respectively.

## Electromagnetic Compatibility (EMC)

A metal oxide VDR together with an input fuse and an input filter form an effective protection against high input transient voltages which typically occur in most installations. The $S$ series has been successfully tested to the following specifications:

## Electromagnetic Immunity

Table 8: Immunity type tests

| Phenomenon | Standard ${ }^{1}$ | Level | Coupling mode ${ }^{2}$ | Value applied | Waveform | Source imped. | Test procedure | In oper. | Perform. ${ }^{3}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Voltage surge | IEC 60571-1 |  | i/c, +i/-i | $800 \mathrm{~V}_{\mathrm{p}}$ | $100 \mu \mathrm{~s}$ | $100 \Omega$ | 1 pos. and 1 neg. voltage surge per coupling mode | yes | 4 |
|  |  |  |  | $1500 \mathrm{~V}_{\mathrm{p}}$ | $50 \mu \mathrm{~s}$ |  |  |  |  |
|  |  |  |  | $3000 \mathrm{~V}_{\mathrm{p}}$ | $5 \mu \mathrm{~s}$ |  |  |  |  |
|  |  |  |  | $4000 \mathrm{~V}_{\mathrm{p}}$ | $1 \mu \mathrm{~s}$ |  |  |  |  |
|  |  |  |  | $7000 \mathrm{~V}_{\mathrm{p}}$ | 100 ns |  |  |  |  |
| Supply related surge | RIA 12 | B | +i/-i | 1.5 • $U_{\text {batt }}$ | 1 s | $0.2 \Omega$ | 1 positive surge | yes | 4 |
| Direct transient |  | C | +i/c, -i/c | $960 \mathrm{~V}_{\mathrm{p}}$ | 10/100 $\mu \mathrm{s}$ | $5 \Omega$ | 5 pos. and 5 neg. impulses | yes | 4 |
|  |  | D |  | $1800 \mathrm{~V}_{\mathrm{p}}$ | 5/50 $\mu \mathrm{s}$ |  |  |  |  |
|  |  | E |  | $3600 \mathrm{~V}_{\mathrm{p}}$ | 0.5/5 $\mu \mathrm{s}$ | $100 \Omega$ |  |  |  |
|  |  | F |  | $4800 \mathrm{~V}_{\mathrm{p}}$ | 0.1/1 $\mu \mathrm{s}$ |  |  |  |  |
|  |  | G |  | $8400 \mathrm{~V}_{\mathrm{p}}$ | 0.05/0.1 $\mu \mathrm{s}$ |  |  |  |  |
| Indirect coupled transient |  | H | -o/c, +o/c | $1800 \mathrm{~V}_{\mathrm{p}}$ | 5/50 $\mu \mathrm{s}$ |  |  |  |  |
|  |  | J |  | $3600 \mathrm{~V}_{\mathrm{p}}$ | 0.5/5 $\mu \mathrm{s}$ |  |  |  |  |
|  |  | K |  | $4800 \mathrm{~V}_{\mathrm{p}}$ | 0.1/1 $\mu \mathrm{s}$ |  |  |  |  |
|  |  | L |  | $8400 \mathrm{~V}_{\mathrm{p}}$ | 0.05/0.1 $\mu \mathrm{s}$ |  |  |  |  |
| Electrostatic discharge (to case) | $\begin{aligned} & \text { IEC/EN } \\ & 61000-4-2 \end{aligned}$ | 4 | contact discharge | $8000 \mathrm{~V}_{\mathrm{p}}$ | $1 / 50 \mathrm{~ns}$ | $330 \Omega$ | 10 positive and 10 negative discharges | yes | A |
|  |  |  | air discharge | $10 \mathrm{~V} / \mathrm{m}$ |  |  |  |  |  |
| Electromagnetic field | $\begin{aligned} & \text { IEC/EN } \\ & 61000-4-3 \end{aligned}$ | 3 | antenna | $10 \mathrm{~V} / \mathrm{m}$ | $\begin{gathered} \text { AM } 80 \% \\ 1 \mathrm{kHz} \end{gathered}$ | n.a. | $80 . . .1000 \mathrm{MHz}$ | yes | A |
| Electromagnetic field, pulse modulated | ENV 50204 |  |  |  | 50\% duty cycle, 200 Hz repetition frequency |  | $900 \pm 5 \mathrm{MHz}$ | yes | A |
| Electrical fast transient/burst | $\begin{aligned} & \text { IEC/EN } \\ & 61000-4-4 \end{aligned}$ | 4 | capacitive, o/c | $2000 \mathrm{~V}_{\mathrm{p}}$ | bursts of $5 / 50 \mathrm{~ns}$ $2.5 / 5 \mathrm{kHz}$ over 15 ms ; burst period: 300 ms | $50 \Omega$ | 1 min positive 1 min negative transients per coupling mode | yes | A |
|  |  |  | $\mathrm{i} / \mathrm{c},+\mathrm{i} /-\mathrm{i}$ direct | $4000 \mathrm{~V}_{\mathrm{p}}$ |  |  |  |  | A |
| Surge | $\begin{aligned} & \text { IEC/EN } \\ & 61000-4-5 \end{aligned}$ | 3 | i/c | 2000 Vp | 1.2/50 $\mu \mathrm{s}$ | $12 \Omega$ | 5 pos. and 5 neg. surges per | yes | A |
|  |  | 4 | +i/-i |  |  | $2 \Omega$ |  |  |  |
|  |  |  | $\mathrm{i} / \mathrm{c}, \mathrm{l} / \mathrm{l}-\mathrm{i}$ | $2500 \mathrm{~V}_{\mathrm{p}}$ | 10/700 $\mu \mathrm{s}$ | $40 \Omega$ |  |  |  |
| Conducted disturbances | $\begin{aligned} & \text { IEC/EN } \\ & 61000-4-6 \end{aligned}$ | 3 | i, o, signal wires | $\begin{gathered} 10 \mathrm{~V}_{\mathrm{rms}} \\ (140 \mathrm{~dB} \mu \mathrm{~V}) \end{gathered}$ | AM 80\% <br> 1 kHz | $150 \Omega$ | $0.15 . . .80 \mathrm{MHz}$ | yes | A |

${ }^{1}$ Related and previous standards are referenced in: Technical Information: Standards.
${ }^{2} \mathrm{i}=$ input, $\mathrm{o}=$ output, $\mathrm{c}=$ case.
${ }^{3} A=$ Normal operation, no deviation from specifications, $B=$ Normal operation, temporary deviation from specs possible
${ }^{4}$ Test in progress, please consult factory.
Note: Previous standards are referenced in: Technical In-
formation: Standards

## Electromagnetic Emission



Fig. 23
Typical disturbance voltage (quasi-peak) at the input according to CISPR 11/22 and EN 55011/22, measured at $U_{\text {inom }}$ and $I_{0}$ nom.


Fig. 24
Typical radiated electromagnetic field strength (quasipeak) according to CISPR 11/22 and EN 55011/22, normalized to a distance of 10 m , measured at $U_{\mathrm{i} \text { nom }}$ and Io nom.

## Immunity to Environmental Conditions

Table 9: Environment specifications

| Test method |  | Standard | Test conditions |  | Status |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Ca | Damp heat steady state | IEC/DIN IEC 60068-2-3 <br> MIL-STD-810D section 507.2 | Temperature: Relative humidity: Duration: | $\begin{aligned} & 40^{ \pm 2}{ }^{\circ} \mathrm{C} \\ & 93^{+2 /-3 \%} \\ & 56 \text { days } \\ & \hline \end{aligned}$ | Unit not operating |
| Ea | Shock (half-sinusoidal) | IEC/EN/DIN EN 60068-2-27 MIL-STD-810D section 516.3 | Acceleration amplitude: Bump duration: Number of bumps: | $\begin{aligned} & 100 \mathrm{~g}_{\mathrm{n}}=981 \mathrm{~m} / \mathrm{s}^{2} \\ & 6 \mathrm{~ms} \\ & 18 \text { (3 each direction) } \end{aligned}$ | Unit operating |
| Eb | Bump (half-sinusoidal) | IEC/EN/DIN EN 60068-2-29 MIL-STD-810D section 516.3 | Acceleration amplitude: Bump duration: Number of bumps: | ```40 gn= 392 m/s}\mp@subsup{}{}{2 ms 6 0 0 0 ~ ( 1 0 0 0 ~ e a c h ~ d i r e c t i o n )``` | Unit operating |
| Fc | Vibration (sinusoidal) | IEC/EN/DIN EN 60068-2-6 | Acceleration amplitude: <br> Frequency (1 Oct/min): <br> Test duration: | $\begin{aligned} & 0.35 \mathrm{~mm}(10 \ldots 60 \mathrm{~Hz}) \\ & 5 \mathrm{~g}_{\mathrm{n}}=49 \mathrm{~m} / \mathrm{s}^{2}(60 \ldots 2000 \mathrm{~Hz}) \\ & 10 \ldots 2000 \mathrm{~Hz} \\ & 7.5 \mathrm{~h}(2.5 \mathrm{~h} \text { each axis }) \end{aligned}$ | Unit operating |
| Fn | Vibration broad band random (digital dontrol) | IEC 60068-2-64 <br> DIN 40046 part 23 <br> MIL-STD-810D section 514.3 | Acceleration spectral density: <br> Frequency band: <br> Acceleration magnitude: <br> Test duration: | $\begin{aligned} & 0.05 \mathrm{gn}^{2} / \mathrm{Hz} \\ & 5 \ldots . .500 \mathrm{~Hz} \\ & 4.97 \mathrm{~g}_{\mathrm{n} \text { rms }} \\ & 3 \mathrm{~h}(1 \mathrm{~h} \text { each axis }) \end{aligned}$ | Unit operating |
| Kb | Salt mist, cyclic (sodium chloride NaCl solution) | IEC/EN/DIN IEC 60068-2-52 | Concentration: <br> Duration: <br> Storage: <br> Storage duration: <br> Number of cycles: | $5 \%\left(30^{\circ} \mathrm{C}\right)$ <br> 2 h per cycle $40^{\circ} \mathrm{C}, 93 \%$ rel. humidity 22 h per cycle 3 | Unit not operating |

Table 10: Temperature specifications, values given are for an air pressure of $800 \ldots 1200 \mathrm{hPa}$ ( $800 \ldots 1200 \mathrm{mbar}$ )

| Temperature |  |  | Standard -7 |  | Option -9 |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Characteristics |  | Conditions | min | max | min | max |  |
| $T_{\text {A }}$ | Ambient temperature | $U_{i \text { min }} \ldots U_{\text {i max }}$ | -25 | 71 | -40 | 71 | ${ }^{\circ} \mathrm{C}$ |
| $T_{\text {C }}$ | Case temperature | $I_{0}=0 . . . I_{\text {o nom }}$ | -25 | 95 | -40 | 95 |  |
| $T_{\text {S }}$ | Storage temperature | Not operational | -40 | 100 | -55 | 100 |  |

Table 11: MTBF

| Values at Specified <br> Case Temperature | Type | Ground Benign | Ground Fixed |  | Ground Mobile | Unit |
| :--- | :--- | :---: | :---: | :---: | :---: | :---: |
| MTBF $^{1}$ |  | $40^{\circ} \mathrm{C}$ | $40^{\circ} \mathrm{C}$ | $70^{\circ} \mathrm{C}$ | $50^{\circ} \mathrm{C}$ |  |

${ }^{1}$ Calcualted in accordance with MIL-HDBK217F.

## Mechanical Data

Dimensions in mm . Tolerances $\pm 0.3 \mathrm{~mm}$ unless otherwise indicated.




Back plate

## Note:

- $\mathrm{d} \geq 15 \mathrm{~mm}$, recommended minimum distance to next part to ensure proper air circulation at full output power.
- free air locations: the module should be mounted with fins in vertical position to achieve a maximum air flow through the heat sink.
Fig. 25
Aluminium case S02 with heatsink, black finish and self cooling, weight: Approx. 1.25 kg


Fig. 26
Aluminium case S02 with option B1 (cooling plate), black finish and self cooling. Total weight: Approx. 1.15 kg


Note: Long case with option B2, elongated by 60 mm for 220 mm rack depth, is available on request. (No LEDs, no test jacks.)

## Safety and Installation Instructions

## Connector Pin Allocation

The connector pin allocation table defines the electrical potentials and the physical pin positions on the H 15 connector. Pin no. 24, the protective earth pin present on all LS AC-DC converters is leading, ensuring that it makes contact with the female connector first.


Fig. 27
View of module's male H15 connector
Table 12: H15 Connector pin allocation

| Pin | Conn | r type H 15 |  |  |
| :---: | :---: | :---: | :---: | :---: |
| No. | LS 40 |  | LS 50 |  |
| 4 | Vo1+ |  | Vo2+ | Output |
| 6 | Vo1+ |  | Vo2+ |  |
| 8 | Vo1- |  | Vo2- |  |
| 10 | Vo1- |  | Vo2- |  |
| 12 | S+ | Sense | Vo1+ | Output 1 |
| 14 | S- | Sense | Vo1- | Output 1 |
| 16 | $\mathrm{R}^{1}$ | Control of $U_{01}$ | R ${ }^{1}$ | Control of $U_{01}$ |
| 18 | i | Inhibit | i | Inhibit |
| 20 | D | Save data | D | Save data |
|  | $\mathrm{V}^{3}$ | ACFAIL |  |  |
| 22 | T | Current sharing | T | Current sharing |
| $24{ }^{2}$ | 방 | Protective earth | $\dagger$ | Protective earth |
| 26 | N | Neutral | $\mathrm{N} \sim$ | Neutral |
| 28 | N~ |  | N~ |  |
| 30 | P~ | Phase | P~ | Phase |
| 32 | $\mathrm{P} \sim$ |  | $\mathrm{P} \sim$ |  |

${ }^{1}$ Feature $R$ excludes option $P$ and vice versa
${ }^{2}$ Leading pin (pregrounding)
${ }^{3}$ Option D excludes option V and vice versa

## Protection Degree

Condition: Female connector fitted to the unit.
IP 30: All units except those with option P, and except those with option D or V with potentiometer.
IP 20: All units fitted with option P , or with option D or V with potentiometer.

## Installation Instructions

The S series AC-DC converters are components, intended exclusively for inclusion within other equipment by an industrial assembly operation or by professional installers. Installation must strictly follow the national safety regulations in compliance with the enclosure, mounting, creepage, clearance, casualty, markings and segregation requirements of the end-use application.
Connection to the system shall be made via the female connector H15 (see: Accessories). Other installation methods may not meet the safety requirements.
The AC-DC converters are provided with pin no. $24(\Theta)$, which is reliably connected with their case. For safety reasons it is essential to connect this pin with the protective earth of the supply system.
An input fuse is built-in in the connection from pins no. 30 and 32 ( $\mathrm{P} \sim$ ) of the unit. Since this fuse is designed to protect the unit in case of an overcurrent and does not necessarily cover all customer needs, an external fuse suitable for the application and in compliance with the local requirements might be necessary in the wiring to one or both input potentials, pins nos. 26 and 28 and/or nos. 30 and 32.

Important: Whenever the inhibit function is not in use, pin no. 18 (i) should be connected to pin no. 14 (S-/Vo1-) to enable the output(s).
Do not open the modules, or guarantee will be invalidated.

Due to high current values, all LS units provide two internally parallel connected contacts for certain paths (pins 4/6, $8 / 10,26 / 28$ and $30 / 32$, respectively). It is recommended to connect load and supply to both female connector pins of each path in order to keep the voltage drop across the connector pins to an absolute minimum and to not overstress the connector contacts if currents are higher than approx. 8 A . The connector contacts are rated 8 A over the whole temperature range.
Make sure that there is sufficient air flow available for convection cooling. This should be verified by measuring the case temperature when the unit is installed and operated in the end-use application. The maximum specified case temperature $T_{\mathrm{C} \text { max }}$ shall not be exceeded. See also: Thermal Considerations.
Check for hazardous voltages before altering any connections.

Ensure that a unit failure (e.g. by an internal short-circuit) does not result in a hazardous condition. See also: Safety of operator accessible output circuit.

## Cleaning Agents

In order to avoid possible damage, any penetration of cleaning fluids is to be prevented, since the power supplies are not hermetically sealed.

## Standards and Approvals

All AC-DC converters correspond to class I equipment. They are UL recognized according to UL 1950, UL recognized for Canada to CAN/CSA C22.2 No. 950-95 and LGA approved to IEC/EN 60950 standards.
The units have been evaluated for:

- Building in
- Basic insulation between input and case, based on 250 V AC and 400 V DC
- Double or reinforced insulation between input and output, based on 250 V AC and 400 V DC
- Basic insulation between output and case based on 200 V AC and DC
- Operational insulation between output and output
- Connecting the input to a primary or secondary circuit which is subject to a maximum transient rating of 2500 V (overvoltage category III based on a 110 V primary circuit, overvoltage category II based on a 230 V primary circuit).
- The use in a pollution degree 2 environment
- The UL 1950 recognition limits the minimum input voltage to $U_{i}=U_{i \min }+5 \mathrm{VAC}=90 \mathrm{VAC}$ if the case temperature exceeds $90^{\circ} \mathrm{C}$.
The AC-DC converters are subject to manufacturing surveillance in accordance with the above mentioned UL, CSA, EN and with ISO 9001 standards.


## Isolation

The electric strength test is performed as factory test in accordance with IEC/EN 60950 and UL 1950 and should not be repeated in the field. Power-One will not honour any guarantee claims resulting from electric strength field tests.

Important: Testing by applying AC voltages will result in high and dangerous leakage currents flowing through the Y-capacitors (see fig.: Block diagram).

Table 13: Isolation

| Characteristic |  | Input to | Input to | Output to | Output to | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Electric strength test voltage | Required according to IEC/EN 60950 | 1.5 | $3.0{ }^{1}$ | - | - | $\mathrm{kV}_{\text {rms }}$ |
|  |  | 2.1 | $4.2{ }^{1}$ | - | - | kV DC |
|  | Actual factory test 1 s | 2.8 | $5.6{ }^{1}$ | 1.4 | 0.14 |  |
|  | AC test voltage equivalent to actual factory test | 2.0 | $4.0{ }^{1}$ | 1.0 | 0.1 | kV rms |
| Insulation resistance at 500 V DC |  | >300 | >300 | >300 | $>100^{2}$ | $\mathrm{M} \Omega$ |

${ }^{1}$ In accordance with IEC/EN 60950 only subassemblies are tested in factory with this voltage.
${ }^{2}$ Tested at 100 V DC.
For creepage distances and clearances refer to: Technical Information: Safety.

## Leakage Currents in AC-DC operation

Leakage currents flow due to internal leakage capacitance and RFI suppression Y-capacitors. The current values are proportional to the mains voltage and nearly proportional to the mains frequency and are specified at an input voltage of $254 \mathrm{~V}(50 \mathrm{~Hz})$ where phase, neutral and protective earth are correctly connected as required for class I equipment.


Fig. 28
Measuring instrument (MI) for earth leaking current tests according to IEC/EN 60950.

Under test conditions the leakage current flows through a measuring instrument (MI) as described in fig.: Measuring instrument for earth leakage current tests, which takes into account impedance and sensitivity of a person touching unearthed accessible parts. The current value is calculated by dividing the measured voltage by $500 \Omega$. If inputs of $S$ units are connected in parallel, their individual leakage currents are added.


Fig. 29
Test set-up

Table 14: Leakage currents

| Characteristic |  | Class I <br> LS $4000 \ldots$ LS 5000 | Unit |
| :--- | :--- | :---: | :---: |
| Maximum earth <br> leakage current | Permissible according to IEC/EN 60950 | 3.5 | mA |
|  | Specified value at $254 \mathrm{~V}, 50 \mathrm{~Hz}$ | 0.82 |  |

## Safety of operator accessible output circuit

If the output circuit of an AC-DC converter is operator accessible, it shall be an SELV circuit according to the IEC/EN 60950 related safety standards.
The following table shows a possible installation configuration, compliance with which causes the output circuit of an S series AC-DC converter to be an SELV circuit according
to IEC/EN 60950 up to a configured output voltage (sum of nominal voltages if in series or +/- configuration) of 36 V .
However, it is the sole responsibility of the installer to assure the compliance with the relevant and applicable safety regulations. More information is given in: Technical Information: Safety.

Table 15: Safety concept leading to an SELV output circuit

| Conditions | AC-DC converter | Installation | Result |
| :--- | :--- | :--- | :--- |
| Nominal voltage | Grade of insulation <br> between input and output <br> provided by the AC-DC converter | Measures to achieve the resulting <br> safety status of the output circuit | Safety status of the AC-DC <br> converter output circuit |
| Mains <br> $\leq 250 ~ V ~ A C ~$ | Double or reinforced | Earthed case ${ }^{1}$ and installation <br> according to the applicable standards | SELV circuit |

${ }^{1}$ The earth connection has to be provided by the installer according to the relevant safety standards, e.g. IEC/EN 60950.


Fig. 30
Schematic safety concept. Use fuses and earth connection as per Installation Instructions and table Safety concept leading to an SELV output circuit.

## Description of Options

Table 16: Survey of options

| Option | Function of option | Characteristic |
| :---: | :--- | :--- |
| -9 | Extended operational ambient temperature range | $T_{\mathrm{A}}=-40 \ldots 71^{\circ} \mathrm{C}$ |
| E | Electronic inrush current limitation circuitry | Active inrush current limitation |
| $\mathrm{P}^{1}$ | Potentiometer for fine adjustment of output voltage | Adjustment range $+10 /-60 \%$ of $U_{\text {o nom, }}$, excludes R input |
| $\mathrm{D}^{2}$ | Input and/or output undervoltage monitoring circuitry | Safe data signal output (D0...DD) |
| $\mathrm{V}^{23}$ | Input and/or output undervoltage monitoring circuitry | ACFAIL signal according to VME specifications (V0, V2, V3) |
| T | Current sharing | Interconnect T-pins if paralleling outputs (5 units max.) |
| B1, B2 | Cooling plate | Replaces standard heat sink, allowing direct chassis-mounting |

${ }^{1}$ Function $R$ excludes option $P$ and vice versa
${ }^{2}$ Option D excludes option V and vice versa
${ }^{3}$ Only available if main output voltage $U_{01}=5.1 \mathrm{~V}$

## -9 Extended Temperature Range

Option -9 extends the operational ambient temperature range from $-25 \ldots 71^{\circ} \mathrm{C}$ (standard) to $-40 \ldots 71^{\circ} \mathrm{C}$. The power supplies provide full nominal output power with convection cooling. Option -9 excludes inrush current limitation by NTC.

## P Potentiometer

The potentiometer provides an output voltage adjustment range of $+10 /-60 \%$ of $U_{0}$ nom and is accessible through a hole in the front cover. This feature enables compensation for voltage drops across the connector and wiring. Option P is not recommended if units are connected in parallel.
Option P excludes the R-function. With double output units both outputs are affected by the potentiometer setting (doubling the voltage setting if the outputs are in series).
If the output voltages are increased above $U_{0}$ nom via R-input control, option $P$ setting, remote sensing or option $T$, the output current(s) should be reduced accordingly so that $P_{\text {o nom }}$ is not exceeded.

## E Inrush Current Limitation

The converters may be supplemented by an electronic circuit (option E, replacing the standard built-in NTC) to achieve an enhanced inrush current limiting function.

Table 17: Inrush current characteristics with option E

| Characteristics$U_{i}=230 \mathrm{VAC}$ |  |  | $\max$ | Unit |
| :---: | :---: | :---: | :---: | :---: |
| $l_{\text {inr }} \mathrm{p}$ | Peak inrush current | - | 25.3 | A |
| $t_{\text {inr }}$ | Inrush current duration | 35 | 50 | ms |



Fig. 31
Option E block diagram


Fig. 32
Inrush current with option $E, U_{\mathrm{i}}=230 \mathrm{~V}$ AC, $P_{\mathrm{o}}=P_{\mathrm{o}}$ nom

## Precaution:

Subsequent switch-on cycles at start-up are limited to max. 10 cycles during the first 20 seconds (cold unit) and at continuing on/off ( $T_{\mathrm{C}}=95^{\circ} \mathrm{C}$ ) max. 1 cycle every 8 sec .

## T Current Sharing

This option ensures that the output currents are approximately shared between all paralleled modules and increases system reliability. To use this facility, simply interconnect the T pins of all modules and make sure, that pin 14, the S- pin (S 4000) or the Vo1- pins (S 5000) are also connected together. The load leads should have equal length and cross section to ensure equal voltage drops. Not more than 5 units should be connected in parallel. If output voltage adjustment is requested we strongly recommend to use the R -input instead of option P , as with option P the required setting accuracy is difficult to achieve. The output voltages must be individually set prior to paralleling to within a tolerance of $1 \ldots 2 \%$ or the R pins should be connected together.


Fig. 33
An example of poor wiring for connection in parallel

max. 5 units connected in parallel
${ }^{1}$ Leads should have equal length and cross sections and should run in the same cable loom.
${ }^{2}$ Diodes recommended in redundant operation only
${ }^{3}$ DC common point
Fig. 34
Paralleling of single output units using option $T$ with the sense lines connected at the load


Fig. 35
Paralleling of double output units using option $T$ with Power bus.

## D Undervoltage Monitor

The input and/or output undervoltage monitoring circuit operates independently of the built-in input undervoltage lockout circuit. A logic "low" (JFET output) or "high" signal (NPN output) is generated at pin 20 as soon as one of the monitored voltages drops below the preselected threshold level $U_{t}$. The return for this signal is Vo1-. The D output recovers when the monitored voltage(s) exceed(s) $U_{t}+U_{\mathrm{h}}$. The
threshold level $U_{t i}$ is adjusted in the factory. The threshold level $U_{\text {to }}$ is either adjusted by a potentiometer, accessible through a hole in the front cover, or factory adjusted to a fixed value specified by the customer.
Option D exists in various versions DO...DD as shown in the following table.

Table 18: Undervoltage monitoring functions

| Output type |  | Monitoring |  | Minimum adjustment range of threshold level $\boldsymbol{U}_{\mathrm{t}}$ |  | Typical hysteresis $U_{\mathrm{ho}}$ [\% of $U_{t}$ ] for $\boldsymbol{U}_{\mathbf{t} \text { min }} \ldots \boldsymbol{U}_{\mathrm{t} \text { max }}$ $U_{\text {ho }}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| JFET | NPN | $\boldsymbol{U}_{\mathbf{i}}$ | $U_{01}$ |  |  |  |
| D1 | D5 | no | yes | - | 3.5... $40 \mathrm{~V}^{1}$ | 2.5...0.6 |
| D2 | D6 | yes | no | 355 V DC ${ }^{4}$ | - | - |
| D3 | D7 | yes | yes | 355V DC ${ }^{4}$ | $\left(0.95 \ldots 0.985 U_{01}\right)^{2}$ | "0" |
| D4 | D8 | no | yes | - | $\left(0.95 \ldots 0.985 U_{01}\right)^{2}$ | "0" |
| D0 | D9 | no | yes | - | $3.5 \ldots 40 \mathrm{~V}^{3}$ | 2.5...0.6 |
|  |  | yes | yes | 355 V DC ${ }^{4}$ | $3.5 . . .40 \mathrm{~V}^{3}$ | 2.5...0.6 |
|  | DD | yes | yes | 355 V DC ${ }^{4}$ | 3.5... $40 \mathrm{~V}^{1}$ | 2.5...0.6 |

${ }^{1}$ Threshold level adjustable by potentiometer
${ }^{2}$ Fixed value. Tracking if $U_{01}$ adjusted via $R$-input, option P or sense lines.
${ }^{3}$ The threshold level permanently adjusted according to customer specification $\pm 2 \%$ at $25^{\circ} \mathrm{C}$. Any value within the specified range is basically possible but causes a special type designation in addition to the standard option designations (D0/D9)!
${ }^{4}$ Option D monitors the boost regulator output voltage. The trigger level is adjusted in the factory to 355 V DC.

JFET output (D0...D4):
Connector pin D is internally connected via the drainsource path of a JFET (self-conducting type) to the negative potential of output 1. $U_{D} \leq 0.4 \mathrm{~V}$ (logic low) corresponds to a monitored voltage level ( $U_{i}$ and/or $U_{01}$ ) $<U_{t}$. The current $I_{D}$ through the JFET should not exceed 2.5 mA . The JFET is protected by a 0.5 W Zener diode of 8.2 V against external overvoltages.

| $U_{\mathrm{i}}, U_{01}$ status | D output, $U_{\mathrm{D}}$ |
| :---: | :---: |
| $U_{\mathrm{i}}$ or $U_{01}<U_{\mathrm{t}}$ | low, $\mathrm{L}, U_{\mathrm{D}} \leq 0.4 \mathrm{~V}$ at $I_{\mathrm{D}}=2.5 \mathrm{~mA}$ |
| $U_{\mathrm{i}}$ and $U_{01}>U_{\mathrm{t}}+U_{\mathrm{h}}$ | high, $\mathrm{H}, I_{\mathrm{D}} \leq 25 \mu \mathrm{~A}$ at $U_{\mathrm{D}}=5.25 \mathrm{~V}$ |

NPN output (D5...DD):
Connector pin D is internally connected via the collectoremitter path of a NPN transistor to the negative potential of output 1. $U_{D}<0.4 \mathrm{~V}$ (logic low) corresponds to a monitored voltage level $\left(U_{i}\right.$ and/or $\left.U_{01}\right)>U_{t}+U_{h}$. The current $I_{D}$ through the open collector should not exceed 20 mA . The NPN output is not protected against external overvoltages. $U_{D}$ should not exceed 40 V .

| $U_{\mathrm{i}}, U_{01}$ status | D output, $U_{\mathrm{D}}$ |
| :---: | :---: |
| $U_{\mathrm{i}}$ or $U_{01}<U_{\mathrm{t}}$ | high, $\mathrm{H}, I_{\mathrm{D}} \leq 25 \mu \mathrm{~A}$ at $U_{\mathrm{D}}=40 \mathrm{~V}$ |
| $U_{\mathrm{i}}$ and $U_{01}>U_{\mathrm{t}}+U_{\mathrm{h}}$ | low, $\mathrm{L}, U_{\mathrm{D}} \leq 0.4 \mathrm{~V}$ at $I_{\mathrm{D}}=20 \mathrm{~mA}$ |



Fig. 36
Option D0...D4: JFET output, $I_{D} \leq 2.5 \mathrm{~mA}$


Fig. 37
Option D5...DD: NPN output, $U_{01} \leq 40 \mathrm{~V}, \mathrm{I}_{\mathrm{D}} \leq 20 \mathrm{~mA}$

Table 19: D-output logic signals

| Version of D | $\boldsymbol{U}_{\mathbf{i}}<\boldsymbol{U}_{\mathbf{t}}$ resp. $\boldsymbol{U}_{\mathbf{0}}<\boldsymbol{U}_{\mathbf{t}}$ | $\boldsymbol{U}_{\mathbf{i}}>\boldsymbol{U}_{\mathbf{t}}+\boldsymbol{U}_{\mathbf{h}}$ resp. $\boldsymbol{U}_{\mathbf{0}}>\boldsymbol{U}_{\mathbf{t}}$ | Configuration |
| :--- | :---: | :---: | :---: |
| D1, D2, D3, D4, D0 | low | high | JFET |
| D5, D6, D7, D8, D9, DD | high | low | NPN |

D-signal with respect to input and output voltage versus time:

## Input voltage monitoring



Output voltage monitoring

${ }^{1}$ Hold-up time see section Electrical Input Data.
2 With output voltage monitoring, hold-up time $t_{\mathrm{h}}=0$.
${ }^{3}$ The signal will remain high if the D output is connected to an external source.
${ }^{4} t_{\text {low } \min }=100 \ldots 170 \mathrm{~ms}$, typically 130 ms .

Fig. 38
Relationship between $U_{\mathrm{ci}}, U_{01}, U_{\mathrm{D}}, U_{01} / U_{0 \text { nom }}$ versus time

## V ACFAIL Signal (VME)

Available for units with $U_{01}=5.1 \mathrm{~V}$
This option defines an undervoltage monitoring circuit for the input or input and main output voltage equivalent to option D and generates the ACFAIL signal (V signal) which conforms to the VME standard.
The low state level of the ACFAIL signal is specified at a sink current of $I_{V} \leq 48 \mathrm{~mA}$ to $U_{V} \leq 0.6 \mathrm{~V}$ (open-collector output of a NPN transistor). The pull-up resistor feeding the open-collector output should be placed on the VME back plane.
After the ACFAIL signal has gone low, the VME standard requires a hold-up time $t_{h}$ of at least 4 ms before the 5.1 V output drops to 4.875 V when the output is fully loaded. This hold-up time $t_{\mathrm{h}}$ is provided by the internal input capacitance. See also fig.: Hold-up Time versus Output Power.

Table 20: Undervoltage monitor functions

| V output |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| (VME compatible) | Monitoring <br> $\boldsymbol{U}_{\mathbf{i}}$ |  | Minimum adjustment <br> range of threshold level <br> $U_{\mathrm{ti}}$ |  |
| V2 | yes | no | 355 V DC ${ }^{1}$ | - |
| V3 | yes | yes | 355 V DC ${ }^{1}$ | $0.95 \ldots 0.985 U_{01}{ }^{2}$ |

${ }^{1}$ Option D monitors the boost regulator output voltage. The trigger level is adjusted in the factory to 355 V DC.
${ }^{2}$ Fixed value between $95 \%$ and $98.5 \%$ of $U_{01}$.
Option V operates independently of the built-in input undervoltage lock-out circuit. A logic "low" signal is generated at pin 20 as soon as one of the monitored voltages drops below the preselected threshold level $U_{t}$. The return for this signal is Vo1-. The V output recovers when the monitored voltage(s) exceed(s) $U_{t}+U_{\mathrm{h}}$. The threshold level $U_{\mathrm{ti}}$ is adjusted in the factory to 355 V DC. The threshold level $U_{\text {to }}$ either is adjusted during manufacture to a determined customer specified value.

V output (V2, V3):
Connector pin V is internally connected to the open collector of a NPN transistor. The emitter is connected to the negative potential of output $1 . U_{\mathrm{V}} \leq 0.6 \mathrm{~V}$ (logic low) corresponds to a monitored voltage level ( $U_{i}$ and/or $U_{01}$ ) $<U_{t}$. The current $I_{V}$ through the open collector should not exceed 50 mA . The NPN output is not protected against external overvoltages. $U_{V}$ should not exceed 60 V .

| $U_{\mathrm{i}}, U_{01}$ status | V output, $U_{\mathrm{V}}$ |
| :---: | :---: |
| $U_{\mathrm{i}}$ or $U_{01}<U_{\mathrm{t}}$ | low, $\mathrm{L}, U_{\mathrm{V}} \leq 0.6 \mathrm{~V}$ at $I_{\mathrm{V}}=50 \mathrm{~mA}$ |
| $U_{\mathrm{i}}$ and $U_{01}>U_{\mathrm{t}}+U_{\mathrm{h}}$ | high, $\mathrm{H}, I_{\mathrm{V}} \leq 25 \mu \mathrm{~A}$ at $U_{\mathrm{V}}=5.1 \mathrm{~V}$ |



Fig. 39
Output configuration of options V2 and V3



${ }^{1}$ VME request: minimum 4 ms
${ }^{2} t_{\text {low min }}=40 \ldots 200 \mathrm{~ms}$, typically 80 ms


${ }^{3} U_{V}$ level not defined at $U_{01}<2.0 \mathrm{~V}$
${ }^{4}$ The V signal drops simultaneously with the output voltage if the pull-up resistor $R_{\mathrm{P}}$ is connected to Vo1+.
The V signal remains high if $R_{\mathrm{P}}$ is connected to an external source.

Output voltage failure
Fig. 40
Relationship between $U_{\mathrm{ci}}, U_{01}, U_{\mathrm{V}}, I_{\mathrm{v}}$ and $U_{01} / U_{\mathrm{o}}$ nom versus time.

## B1 Cooling Plate (see: Mechanical Data)

Where a cooling surface is available, we recommend the use of a cooling plate (option B1) instead of the standard heatsink. The mounting system should ensure sufficient cooling capacity to guarantee that the maximum case temperature $T_{\mathrm{C} \text { max }}$ is not exceeded. The cooling capacity is calculated by:

## Accessories

A variety of electrical and mechanical accessories are available including:

- Front panels for 19" rack mounting, Schroff and Intermas systems.
- Mating H15 connectors with screw, solder, fast-on or press-fit terminals.
- Connector retention facilities.
- Code key system for connector coding.
- Chassis mounting plates for mounting the 19" cassette to a chassis/wall where only frontal access is given.
- Universal mounting bracket for DIN-rail or chassis mounting.
For more detailed information please refer to: Accessory Products.


H15 female connector,
Code key system


Mounting plate,
Connector retention clips
$P_{\text {Loss }}=\frac{(100 \%-\eta)}{\eta}\left(U_{0} \cdot I_{0}\right)$
Efficiency $\eta$ see: Type survey.
Elongated case for 220 mm rack depth: Option B2.


Front panels


Chassis mounting bracket S


Universal mounting bracket for DIN-rail mounting.

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[^0]:    ${ }^{1}$ First column: single output units or double output units with separated outputs, second column: outputs in series connection

