PET2000-12-074xA AC-DC Front-End Power Supply

PET2000-12-074xA is a 2000 Watt AC to DC, power-factor corrected (PFC) power supply that converts standard AC power into a main output of +12 VDC.

PET2000-12-074xA utilizes full digital control architecture for greater efficiency, control and functionality.

This power supply meets international safety standards and displays the CE-Mark for the European Low Voltage Directive (LVD).

Key Features & Benefits

- Best-in-class, 80 PLUS Certified "Platinum" Efficiency
- Auto-Selected Input Voltage Ranges: 90 140 VAC, 180 264 VAC
- AC Input with Power Factor Correction
- 2000 W Continuous Output Power Capability
- Always-On 12 V Standby Output
- Hot-Plug Capable
- Parallel Operation with Active Current Sharing
- Full Digital Controls for Improved Performance
- High Density Design: 42.1 W/in³
- Small Form Factor: 265 x 73.5 x 40 mm (10.43 x 2.89 x 1.57 in)
- Power Management Bus Communication Protocol for Control, Programming and Monitoring
- Status LED with Fault Signaling

Applications

- Networking Switches
- Servers & Routers
- Telecommunications







1. ORDERING INFORMATION

PET	2000	-	12	-	074	x	А	
Product Family	Power Level	Dash	V1 Output	Dash	Width	Airflow	Input	AC Inlet ¹
PET Front-Ends	2000 W		12 V		74 mm	N: Normal R: Reverse	A: AC	Blank: C14 C: C16 A: Saf-D-Grid®

2. OVERVIEW

The PET2000-12-074xA AC/DC power supply is a fully DSP controlled, highly efficient front-end power supply. It incorporates resonance-soft-switching technology to reduce component stresses, providing increased system reliability and very high efficiency. With a wide input operational voltage range the PET2000-12-074xA maximizes power availability in demanding server, network, and other high availability applications. The supply is fan cooled and ideally suited for integration with a matching airflow path.

The PFC stage is digitally controlled using a state-of-the-art digital signal processing algorithm to guarantee best efficiency and unity power factor over a wide operating range.

The DC/DC stage uses soft switching resonant techniques in conjunction with synchronous rectification. An active OR-ing device on the output ensures no reverse load current and renders the supply ideally suited for operation in redundant power systems.

The always-on standby output provides power to external power distribution and management controllers. It is protected with an active OR-ing device for maximum reliability.

Status information is provided with a front-panel LED. In addition, the power supply can be controlled and the fan speed set via the I2C bus. The I2C bus allows full monitoring of the supply, including input and output voltage, current, power, and inside temperatures. Cooling is managed by a fan controlled by the DSP controller. The fan speed is adjusted automatically depending on the actual power demand and supply temperature and can be overridden through the I²C bus.

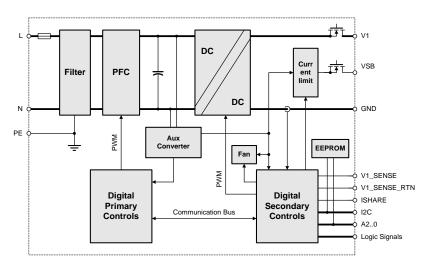


Figure 1. PET2000-12-074xA Block Diagram

3. ABSOLUTE MAXIMUM RATINGS

Stresses in excess of the absolute maximum ratings may cause performance degradation, adversely affect long-term reliability and cause permanent damage to the supply.

PARAMETER		CONDITIONS / DESCRIPTION	MIN	MAX	UNITS
Vi maxc	Maximum Input	Continuous		264	VAC

¹ C14 = IEC 60320-C14 type, C16 = IEC 60320-C16 type, Saf-D-Grid® = Anderson Saf-D-Grid®



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4. INPUT

General Condition: $T_A = 0...55$ °C, unless otherwise noted.

PARAME	ſER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
Vinom	Nominal Input Voltage	Rated Voltage High Line (Vinom HL)	200	230	240	VAC
Vinom	Nominal input voltage	Rated Voltage Low Line (Vinom LL)	100	115	127	VAC
Vi	Input Voltage Ranges	Normal operating (Vi min HL to Vi max HL), High Line	180		264	VAC
Vi	input voltage hanges	Normal operating (Vi min LL to Vi max LL), Low Line	90		140	VAC
		<i>V</i> _{<i>i</i>} =100 VAC, <i>I</i> ₁ = 83 A, <i>I</i> _{SB} = 5 A			13	
		<i>V</i> _{<i>i</i>} = 200 VAC, <i>I</i> _{<i>i</i>} =167 A, <i>I</i> _{<i>SB</i>} = 5 A			12	
li max	Maximum Input Current	<i>V</i> _{<i>i</i>} = 200 VAC, <i>I</i> _{<i>i</i>} = 145 A, <i>I</i> _{<i>SB</i>} = 5 A			10	Arms
		<i>V</i> _{<i>i</i>} = 220 VAC, <i>I</i> _{<i>i</i>} = 158 A, <i>I</i> _{SB} = 5 A			10	
		<i>V</i> _{<i>i</i>} = 230 VAC, <i>I</i> _{<i>i</i>} = 167 A, <i>I</i> _{<i>SB</i>} = 5 A			10	
li inrush	Inrush Current Limitation	$V_{i min}$ to $V_{i max}$, T_{NTC} = 25°C, 5 ms			10	Ap
fi	Input Frequency		47	50/60	63	Hz
	Power Factor	<i>V_i</i> = 230 VAC, 10% load	0.8	0.880		W/VA
PF		<i>V_i</i> = 230 VAC, 20% load	0.9	0.950		W/VA
PF		<i>V</i> _i = 230 VAC, 50% load	0.9	0.997		W/VA
		<i>V</i> _i = 230 VAC, 100% load	0.95	0.999		W/VA
THD	Total Harmonic Distortion	TBD			TBD	%
Vi on	Turn-on Input Voltage ²	Ramping up	87		90	VAC
V i off	Turn-off Input Voltage ²	Ramping down	82		87	VAC
		<i>V</i> _i = 230 VAC, 10% load	90	91.6		%
2	Efficiency ³	<i>V_i</i> = 230 VAC, 20% load	91	93.8		%
η	Enciency	<i>V</i> _i = 230 VAC, 50% load	94	94.4		%
		<i>V</i> /= 230 VAC, 100% load	91	92.8		%
T	Hald up Time I/	V/= 230 VAC, 50% load, 0°	18			ms
Tv1 holdup	Hold-up Time V1	V/= 230 VAC, 100% load, 0°	9			ms
TVSB holdup	Hold-up Time VsB	<i>V</i> /= 90 to 264 VAC, 0 to 100% load	70			ms

4.1 INPUT CONNECTOR

PET2000-12-074NA power supply is available in 3 different input connector configurations. The versions with IEC 60320-C14 and IEC 60320-C16 have a limited current of 10 A for areas outside North America, in addition the IEC 60320-C14 has a limited component temperature of 70°C. The Anderson Saf-D-Grid® has no limitation with respect to both current and temperature.

The PET2000-12-074NA power supply is available with IEC 60320-C14.

Below table shows the maximum rated operating conditions for the different input connector options. The applied operating condition must remain within these conditions to allow safety compliant operation. See also <u>10.3 MAXIMUM OUTPUT POWER VERSUS INLET TEMPERATURE FOR SAFETY COMPLIANCY</u> for detailed derating curves.

² The Front-End is provided with a minimum hysteresis of 3 V during turn-on and turn-off within the ranges

³ Efficiency measured without fan power per EPA server guidelines



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 +1 40

North America +1 408 785 5200 BCD.00478_AL

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ТҮРЕ	INPUT CONNECTOR	REGION	APPLIED RATED MAINS AC VOLTAGE ⁴	MAX /5	MAXIMUM DERATED I ₁ AT MAXIMUM T _A
		North America	100 to 127 VAC	83.3 A	50 A at <i>T</i> _A = 55°C
		North America	200 to 240 VAC	167 A	100 A at <i>T</i> _A = 55°C
PET2000-12-074RA	IEC 60320-C14		100 to 127 VAC	67 A / 83.3 A for BSMI	50 A at <i>T</i> _A = 55°C
		Other than	200 to 220 VAC	145 A	100 A at <i>T</i> _A = 55°C
		North America	220 to 230 VAC	158 A	100 A at <i>T</i> _A = 55°C
			230 to 240 VAC	167 A	100 A at <i>T</i> _A = 55°C
		North America	100 to 127 VAC	83.3 A	50 A at <i>T</i> _A = 70°C
		North America	200 to 240 VAC	167 A	100 A at <i>T</i> _A = 70°C
PET2000-12-074NA IEC	IEC 60320-C14		100 to 127 VAC	67 A / 83.3 A for BSMI	17.5 A at <i>T</i> _A = 65°C
		Other than	200 to 220 VAC	145 A	32.5 A at <i>T</i> _A = 65°C
		North America	220 to 230 VAC	158 A	40 A at <i>T</i> _A = 65°C
			230 to 240 VAC	167 A	43 A at <i>T</i> _A = 65°C
		North America	100 to 127 VAC	83.3 A	50 A at <i>T</i> _A = 70°C
		North America	200 to 240 VAC	167 A	100 A at <i>T</i> _A = 70°C
PET2000-12-074NAC	IEC 60320-C16		100 to 127 VAC	67 A / 83.3 A for BSMI	40 A at $T_A = 70^{\circ}$ C
		Other than	200 to 220 VAC	145 A	87 A at <i>T</i> _A = 70°C
		North America	220 to 230 VAC	158 A	95 A at <i>T</i> _A = 70°C
			230 to 240 VAC	167 A	100 A at <i>T</i> _A = 70°C
PET2000-12-074NAA	Anderson Saf-D-Grid®	All	100 to 127 VAC	83.3 A	50 A at <i>T</i> _A = 70°C
FE12000-12-074NAA	Anderson Sal-D-Grid®	All	200 to 240 VAC	167 A	100 A at <i>T</i> _A = 70°C

4.2 INPUT FUSE

Time-lag 16 A input fuse (5 x 20 mm) in series with the L-line inside the power supply protects against severe defects. The fuse is not accessible from the outside and is therefore not a serviceable part.

4.3 INRUSH CURRENT

The AC-DC power supply exhibits an X-capacitance of only 5.9 µF, resulting in a low and short peak current, when the supply is connected to the mains. The internal bulk capacitor will be charged through an NTC which will limit the inrush current.

NOTE:

Do not repeat plug-in / out operations within a short time, or else the internal in-rush current limiting device (NTC) may not sufficiently cool down and excessive inrush current or component failure(s) may result.

4.4 INPUT UNDER-VOLTAGE

If the sinusoidal input voltage stays below the input undervoltage lockout threshold Vi on, the supply will be inhibited. Once the input voltage returns within the normal operating range, the supply will return to normal operation again.

⁵ Maximum Input current for PET2000-12-074RA at $T_A = 40^{\circ}$ C and for PET2000-12-074NA at $T_A = 55^{\circ}$ C



⁴ Nominal grid voltage, does not include typical fluctuations of ±10%; e.g. listed range 230-240 VAC allows operation

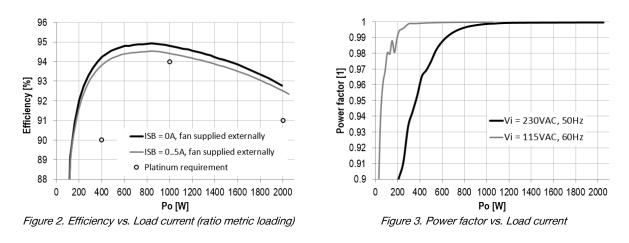
at 230 VAC -10% to 240 VAC +10%, so 207 ... 264 VAC actual voltage to account for grid fluctuations

4.5 POWER FACTOR CORRECTION

Power factor correction (PFC) is achieved by controlling the input current waveform synchronously with the input voltage. A fully digital controller is implemented giving outstanding PFC results over a wide input voltage and load ranges. The input current will follow the shape of the input voltage. If for instance the input voltage has a trapezoidal waveform, then the current will also show a trapezoidal waveform.

4.6 EFFICIENCY

High efficiency (see *Figure 2*) is achieved by using state-of-the-art silicon power devices in conjunction with soft-transition topologies minimizing switching losses and a full digital control scheme. Synchronous rectifiers on the output reduce the losses in the high current output path. The speed of the fan is digitally controlled to keep all components at an optimal operating temperature regardless of the ambient temperature and load conditions.



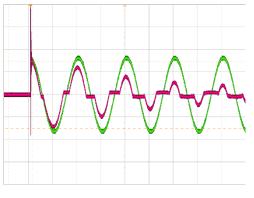


Figure 4. Inrush current, V_i = 230Vac, 90° CH2: V_i (200V/div), CH3: I_i (5A/div)



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5. OUTPUT

5.1 MAIN OUTPUT V1

General Condition: $T_A = 0...40$ °C (PET2000-12-074RA), $T_A = 0...55$ °C (PET2000-12-074NA), Vi = 230 VAC unless otherwise noted.

	PARAME		DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
V_{sec} Output Septoint Accuracy-0.5 40.5 $96 V_{nom}$ $dV_{1 tot}$ Static Regulation $V_{intri I,L}$ to $V_{intri I,K}$ to V_{introm} -1 $+1$ $96 V_{introm}$ $P_{1 nom}$ Nominal Output Power ⁶ $V_{intri I,K}$ to V_{introm} 1000 W $P_{1 nom}$ Peak Output Power ⁶ $V_{intri I,K}$ to V_{introm} 1000 W $P_{1 nom}$ Output Current $V_{intri I,K}$ to V_{introm} 1320 W I_{introm} $V_{intri I,K}$ to V_{introm} 0 167 ADC I_{introm} $V_{intri I,K}$ to V_{introm} 0 175 ADC I_{introm} $V_{intri I,K}$ to V_{introm} 0 175 ADC I_{introm} V_{introm} V_{introm} 0 110 ADC V_{introm} V_{introm} V_{introm} 0 110 ADC V_{introm} V_{introm} V_{introm} 0 110 ADC V_{introm} V_{introm} V_{introm} 120 $mVpp$ V_{introm} V_{introm} 10 100% 110 ADC V_{introm} V_{introm} 10 1050 $mVpp$ V_{introm} V_{introm} 120 $mVpp$ V_{ipp} Output Ripple Voltage ⁸ V_{introm} 120 $mVpp$ V_{ipp} Output Ripple Voltage ⁸ V_{introm} 120 $mVpp$ V_{ipp} Load Regulation V_{introm} 120 $mVpp$ dV_{ipp} Load Regulation </td <td>V_{1 nom}</td> <td>Nominal Output Voltage</td> <td></td> <td></td> <td>12.0</td> <td></td> <td>VDC</td>	V _{1 nom}	Nominal Output Voltage			12.0		VDC
Prime Prime Prime Prime Instruct to Vinaetic Vinnet Li to Vinaetic V	V _{1 set}	Output Setpoint Accuracy	$0.5 \cdot 11 \text{ nom}, 1A = 25^{\circ} \text{C}$	-0.5		+0.5	% V _{1 nom}
PrisonNominal Output PowersWinstatoWinstatoWinstato $P_{I peak}$ Peak Output Powers $WinstattoVinstatta2100WP_{I peak}Peak Output PowersWinstattoVinstatta2100WI nonOutput CurrentWinstattoVinstatta0167ADCI nonMinstattoVinstatta0175ADCI nonMinstattoVinstatta0175ADCI nonMinstattoVinstatta0175ADCI peak indPeak Output Current?Vinstattatto0110ADCVinstattoVinstattattoVinstattatto0110ADCVinstattoVinstattoVinstattatto0110ADCVinstattoVinstattoVinstattatto0110ADCVinstattoVinstattattoVinstattatto0110ADCVinstattoVinstattattoVinstattatto0110ADCVinstattoVinstattattoVinstattatto0110ADCVinstattoVinstattattoVinstattatto0110MVpVinstattoVinstattattoVinstattatto0110MVpVinstattoVinstattattoVinstattatto0100%100%VinstattoVinstattattoVinstattatto0100%100%VinstattoVinstattattoVinstattattoVinstattattoNVPp<$	$dV_{1 tot}$	Static Regulation	$V_{i \min LL}$ to $V_{i \max HL}$, 0 to 100% $I_{1 nom}$	-1		+1	% V _{1 nom}
$ \begin{array}{c c c c c c c c c c c c c c c c c c c $	D,	Nominal Output Powor6	Vimin HL to Vimax HL		2000		W
P_{1peak} Peak Output Power ⁶ $V_{min LL}$ to $V_{max LL}$ 1320W $I_{1non met}$ Output Current $V_{min LL}$ to $V_{max LL}$ 0167ADC $I_{1non met}$ P_{eak} Output Current $V_{min LL}$ to $V_{max LL}$ 0175ADC I_{1peak} P_{eak} Output Current? $V_{min HL}$ to $V_{max LL}$ 0175ADC V_{rpp} P_{eak} Output Current? $V_{min HL}$ to $V_{max LL}$ 0110ADC V_{rpp} Output Ripple Voltage ⁸ $V_{min LL}$ to $V_{max LL}$ 0110MVpp $C_{axe 2}$ or mF $V_{min LL}$ to $V_{max LL}$ 0110MVpp $C_{axe 2}$ or $mF_{LOw LSS}$ $C_{axe 2}$ or $mF_{LOw LSS}$ 120 $mVpp$ dV_{rpp} Output Ripple Voltage ⁸ $C_{axe 2}$ or $mF_{LOw LSS}$ 120 $mVpp$ dV_{rase} Load Regulation $V_{max HL}$ to $V_{max HL}$, 0 to 100% I_{room} -24024 dV_{rase} Load Regulation $V_{max HL}$ to $V_{max HL}$, 0 to 100% I_{room} -24024 dV_{rase} Line Regulation $V_{max HL}$ to $V_{max HL}$, 0 to 100% I_{room} -6+6ADC dV_{rase} Current SharingDifference between individual I_{r} -6+6ADC V_{StatRe} Current SharingDifference between individual I_{r} -6+6ADC V_{StatRe} Current Share Bus Voltage I_{room} R_{room} 0.350.6VDC dV_{rase} Load Transient Response $dA = 50\%$	IInom	Nominal Output I ower-	Vimin LL to Vimax LL		1000		W
$H \operatorname{nom}$ $H \operatorname$	P _{1 pook}	Peak Output Power ⁶	Vimin HL to Vimax HL		2100		W
Output CurrentVimin LL to Vimax LL083ADC h_{poak} h_{poak} $h_{poak ned}$ Peak Output Current?Vimin LL to Vimax LL0110ADC $h_{poak ned}$ $Peak$ Output Current?Vimin LL to Vimax LL0110ADC V_{inpok} $M_{min LL}$ to Vimax LL0110ADC V_{inpok} $M_{min LL}$ to Vimax LL0110ADC V_{inpok} $M_{min LL}$ to Vimax LL0110MVpp V_{inpok} $M_{min LL}$ to Vimax LL0 to 75% h_{inom} 120mVpp V_{inpok} Load Regulation0 to 100% h_{inom} -83-110-138mV dV_i loadLoad Regulation $V_{inhir HL}$ to $V_{imax} H_{i}$, 0.5 · h_{inom} -24024mV/c dV_i loadCurrent Sharing1 8 power supplies in parallel-6+6ADCADC $V_{istAAPE}$ Current Sharing1 8 power supplies in parallel-6+6ADCADC $V_{istAAPE}$ Current Share Bus Voltage $V_{istAAPE at}$ 167A8VDC dV_{ih} Load Transient Response h_i = 00° h_{inom} , h_i = 5 100% h_{inom} 0.350.6VDC dV_{ih} Dynamic Load Regulation dh_i = 60% h_{inom} , h_i = 5 100% h_{inom} 0.51ms dV_{ih} Load Transient Response h_i = 00° h_i 0.10% h_{inom} 0.51ms dV_{ih} Dynamic Load Regulation dh_i = 60% h_{inom} , h_i = 5 100% h_{inom}	i i peak		Vimin LL to Vimax LL		1320		W
<i>h</i> nom red <i>V</i> mm LL to <i>V</i> max LL083ADC <i>h</i> peak <i>h</i> peak Output Current? <i>V</i> mm LL to <i>V</i> max LL0175ADC <i>h</i> peak red <i>V</i> mm LL to <i>V</i> max LL0110ADC <i>V</i> mm LL to <i>V</i> max LL0110ADC <i>V</i> mm LL to <i>V</i> max LL0110ADC <i>V</i> mm LL to <i>V</i> max HL, 0 to 75% <i>h</i> mem, <i>Carr</i> = 0 mF120mVpp <i>V</i> mm LL to <i>V</i> max HL, 0 to 100% <i>h</i> mem, <i>Carr</i> = 0 mF120mVpp <i>dV</i> flageLoad Regulation0 to 100% <i>h</i> mem, <i>Carr</i> = 0 mF120mVpp <i>dV</i> flageLine Regulation <i>V</i> mm LL to <i>V</i> max HL, 0.5 · <i>h</i> nom, <i>Carr</i> = 0 mF-83-110-138mV <i>dV</i> flageLine Regulation <i>V</i> mm HL to <i>V</i> max HL, 0.5 · <i>h</i> nom-24024mV/*C <i>dV</i> flageCurrent SharingDifference between individual <i>h</i> , 1 8 power supplies in parallel-6+6ADC <i>V</i> styleneCurrent Share Bus Voltage <i>h</i> peak9.14VDC <i>dV</i> flageLoad Transient Response <i>h</i> = 50% <i>h</i> nom, <i>h</i> = 5 100% <i>h</i> nom, <i>Carr</i> = 0 mF0.350.6VDC <i>dV</i> flageLoad Transient Response <i>Ah</i> = 50% <i>h</i> nom, <i>h</i> = 5 100% <i>h</i> nom, <i>Carr</i> = 0 mF0.350.6VDC <i>dV</i> flageDynamic Load Regulation <i>f</i> = 50 500% <i>h</i> nom, <i>Carr</i> = 0 mF0.350.6VDC <i>dV</i> flageDun to to 100% <i>h</i> nom, <i>f</i> = 5 100% <i>h</i> nom, <i>Carr</i> = 0 mF0.350.6VDC <i>dV</i> flageRecove	I _{1 nom}	Output Current	Vimin HL to Vimax HL	0		167	ADC
Peak Output Current? $V_{imin LL}$ to $V_{imax LL}$ 0110ADC $h_{ipeak red}$ $V_{imin LL}$ to $V_{imax LL}$ 0110ADC V_{ipp} Output Ripple Voltage8 $V_{imin LL}$ to $V_{max H_0}$ 0 to 75% h_{inom} 120mVpp V_{ipp} Output Ripple Voltage8 $V_{imin LL}$ to $V_{max H_0}$ 75 to 100% h_{inom} 120mVpp dV_i kadLoad Regulation0 to 100% h_{inom} -83-110-138mV dV_i kadLoad Regulation $V_{imin LL}$ to $V_{imax H_0}$ 0.5 · h_{inom} -24024mV'C dV_i tampThermal Drift0.5 · h_{inom} , $T_a = 0 \dots 55^{\circ}$ C-0.4mV'CmV'C dI_i shareCurrent SharingDifference between individual h_i -6+6ADC $V_{SitAdere}$ Current Share Bus Voltage $V_{SitAdere}$ 1 from, $h = 5 \dots$ 100% h_{inom} 0.350.6VDC $dV_{i th}$ Load Transient Response h_{inom} , $h = 0 \dots$ 10% h_{inom} , $n = 0$ 0.350.6VDC $dV_{i th}$ Load Transient Response $\Delta h = 60\%$ h_{inom} , $h = 0 \dots$ 10% h_{inom} 0.51ms $V_{i dyn}$ Dynamic Load Regulation $\Delta h = 60\%$ h_{inom} , $h = 5 \dots 100\%$, h_{inom} 0.51ms $V_{i dyn}$ Dynamic Load Regulation $A h = 60\%$ h_{inom} , $h = 1 \dots .90\%$, h_{inom} 11.412.6V $V_{i dyn}$ Dynamic Load Regulation $A h = 60\%$ h_{inom} , $h = 5 \dots 100\%$ h_{inom} 0.51ms $V_{i dyn}$ Dynamic Load Regulation<	It nom red	output outfolk	Vimin LL to Vimax LL	0		83	ADC
h paak red $V_{min} L_L$ to $V_{max} L_L$ 0110ADC V_{fpp} Output Ripple Voltage8 $V_{min} L_L$ to $V_{max} H_L$, 75 to 100% I_{fnom} , $C_{axt} = 0$ mF120mVpp V_{fpp} $V_{imb} L_L$ to $V_{max} H_L$, 75 to 100% I_{fnom} , $C_{axt} = 0$ mF150mVpp dV_{i} loadLoad Regulation0 to 100% I_{fnom} -83-110-138mV dV_i loadLoad Regulation $V_{imb} H_L$ to $V_{max} H_L$, 0.5 · I_{fnom} -24024mV° dV_i targeThermal Drift0.5 · I_{fnom} , $T_a = 0 55°C-0.4mV°CdI_{i} shareCurrent SharingDifference between individual I_{f_1}\dots spower supplies in parallel-6+6ADCV_{SHARE}Current Share Bus VoltageV_{SHARE at} 167A8VDCV_{SHARE}Load Transient Response\Delta h = 50\% h_{fnom}, h = 5 100\% h_{fnom}, C_{axt} = 0 mF0.350.6VDCdV_{1:h}Load Transient Response\Delta h = 50\% h_{fnom}, h = 5 100\% h_{fnom}, C_{axt} = 0 mF0.350.6VDCdV_{1:h}Load Transient Response\Delta h = 60\% h_{fnom}, h = 0 10\% h_{fnom}, C_{axt} = 0 mF0.350.6VDCdV_{1:h}Dynamic Load RegulationF_{i} = 5 167 A, F_{i} = 5 167 A, F_{i} = 5 100\% h_{fnom}, C_{axt} = 10 90\%, f_{i} = 5 30 mF11.412.6VV_{i dyn}Dynamic Load RegulationF_{i} = 10 90\% h_{fnom}, C_{axt} < 10 mF$	I _{1 peak}	Peak Output Current ⁷	$V_{i \min HL}$ to $V_{i \max HL}$	0		175	ADC
$V_{1,pp}$ Output Ripple Voltage* $C_{oxt} = 0 \text{ mF}$ $V_{innull to V_{inax}H_{il}, 75 to 100% I_{1 nom}$ $C_{oxt} \ge 1 \text{ mFL}$ to $V_{inax}H_{il}, 75 to 100% I_{1 nom}$ $C_{oxt} \ge 1 \text{ mFL}$ $V_{innull to V_{inax}H_{il}, 75 to 100% I_{1 nom}$ $C_{oxt} \ge 1 \text{ mFL}$ $V_{innull to V_{inax}H_{il}, 75 to 100% I_{1 nom}$ $C_{oxt} \ge 1 \text{ mFL}$ $V_{innull to V_{inax}H_{il}, 0 to 100% I_{1 nom}$ $C_{oxt} \ge 1 \text{ mFL}$ $V_{innull to V_{inax}H_{il}, 0.5 \cdot I_{1 nom}$ -24 120mVpp dV_1 ineLine Regulation $V_{inin H_{il}$ to $V_{inax}H_{il}, 0.5 \cdot I_{1 nom}$ -24 -33 -110 -138 mV dV_1 ineDifference between individual I_{i_1} $1 \dots 8 power supplies in parallel-6+6+6ADCV_{SHARE}U_{SHARE}Current SharingDifference between individual I_{i_1}1 \dots 8 power supplies in parallel-6+6+6ADCV_{SHARE}U_{SHARE}Current Share Bus VoltageI_{I poak}9.14VDCV_{VI int}Uoad Transient Response\Delta h = 50\% I_{I nom}, I_I = 5 \dots 100\% I_{I nom}, 0.350.6VDCC_{oxt} = 0 \text{ mF}\Delta h = 10\% I_{I nom}, I_I = 0 \dots 10\% I_{I nom}C_{oxt} = 0 \text{ mF}0.350.6VDCV_{I dyn}Dynamic Load RegulationA_h = 60\% I_{I nom}, I_I = 5 \dots 167 A, I_{I nom}0.51msV_{I dyn}Dynamic Load RegulationA_{I = 10 \dots 30\% V_{I nom}, I_{I = 5 \dots 100\% I_{I nom}0.51msV_{I dyn}Dynamic Load RegulationA_{I = 10,\dots 30\% V_{I nom}, C_{oxt} < 10 \text{ mF}130msV_{I dyn}Dynamic$	I _{1 peak red}			0		110	ADC
$V_{T,pp}$ Output htpple voltage $C_{ext} = 0 \text{ mF}$ $V_{imax}Lt to V_{imax}Lt, 0 to 100% l_{1 nom},C_{ext} \geq 1 \text{ mF/Low ESR}120mVppdV_1 loadLoad Regulation0 to 100% l_{1 nom}-83-110-138mVdV_1 lineLine RegulationV_{imin}Lt to V_{imax}Ht, 0.5 · l_{1 nom}-24024mVdV_1 lineLine RegulationV_{imin}Lt to V_{imax}Ht, 0.5 · l_{1 nom}-24024mVdV_1 lineUnite RegulationV_{imin}Lt to V_{imax}Ht, 0.5 · l_{1 nom}-24024mV'°CdV_1 lineUnite RegulationV_{imin}Lt to V_{imax}Ht, 0.5 · l_{1 nom}-24024mV'°CdV_1 lineUnite RegulationV_{imin}Lt to V_{imax}Ht, 0.5 · l_{1 nom}-24024mV'°CdV_1 lineUniterence between individual l_{r_1}-6+6ADCADCV_{ISHARE}Current Share Bus VoltageV_{ISHARE at 167A8VDCV_{ISHARE}Current Share Bus Voltagel_{1 peak}9.14VDCdV_{1:th}Load Transient Response\Delta h = 50\% l_{1 nom}, l_{1} = 5 \dots 100\% l_{1 nom}, cext = 0 mF0.350.6VDCdV_{1:th}Load Transient Response\Delta h = 60\% l_{1 nom}, l_{1} = 0 \dots 10\% l_{1 nom}, cext = 10 mF0.350.6VDCdV_{1:th}Dynamic Load Regulationf_{1} = 5 \dots 500 Hz, nom, l_{1} = 5 \dots 167 A, l_{2} = 5 \dots 500 Hz, nom, l_{1} = 5 \dots 500 Hz, nom, l_{2} < 10 \text{ mF}130msV_{1:the}$						120	mVpp
$V_{infini} LL$ to $V_{imax} HL$, 0 to 100% $I_{1 nom,}$ 120mVpp $dV_{1 load}$ Load Regulation0 to 100% $I_{1 nom}$ -83-110-138mV $dV_{1 line}$ Line Regulation $V_{imin} HL$ to $V_{imax} HL$, 0.5 · $I_{1 nom}$ -24024mV $dV_{1 line}$ Difference between individual I_{i} -24024mV°C $dI_{1 share}$ Current SharingDifference between individual I_{i} -6+6ADC V_{SHARE} Current Share Bus Voltage V_{SHARE} at 167A8VDC V_{SHARE} Current Share Bus Voltage I_{ipeak} 9.14VDC $dV_{1 fit}$ Load Transient Response $\Delta h = 50\% I_{1 nom}, I_{1} = 0 \dots 10\% I_{1 nom}, C_{ext} = 0 mF$ 0.350.6VDC $dV_{1 fit}$ Dynamic Load Regulation $A_{f} = 60\% I_{1 nom, I_{1}} = 5 \dots 100\% I_{1 nom}, C_{ext} = 10 mF$ 1.30ms $V_{1 dyn}$ Dynamic Load Regulation $A_{f} = 50 \dots 5000 Hz, puty cycle = 10 \dots 90\%, 11.4$ 12.6V $V_{1 dyn}$ Dynamic Load Regulation $V_{f = 50} \dots 5000 Hz, puty cycle = 10 \dots 90\%, 11.4$ 12.6V $V_{1 dyn}$ Output Voltage Rise Time $V_{f = 10 \dots 90\% V_{f nom}, C_{ext} < 10 mF$ 130ms $t_{V1 oris sh}$ Output Turn-on Overshoot0 to 100% $I_{f nom}$ 0.6VV $dV_{f serse}$ Remote SenseCompensation for cable drop, 0 to 100% $I_{f nom}$ 0.25V	V _{1 pp}	Output Ripple Voltage ⁸				150	mVpp
dV_1 lineLine Regulation $V_{imin HL}$ to $V_{imax HL}$, $0.5 \cdot I_{1 nom}$ -24 024mV dV_1 tempThermal Drift $0.5 \cdot I_{1 nom}$, $T_a = 0 \dots 55^{\circ}$ C -0.4 mV/°C dI_1 shareCurrent SharingDifference between individual I_1 , $1 \dots 8$ power supplies in parallel -6 $+6$ ADC V_{ISHARE} Current Share Bus Voltage $V_{ISHARE at 167A$ 8 VDC V_{ISHARE} Current Share Bus Voltage $I_{1 \text{ ceak}}$ 9.14 VDC $dV_{1 It}$ Load Transient Response $Ah = 50\% (1 \text{ nom}, I_1 = 5 \dots 100\% I_1 \text{ nom}, Coxt = 0 \text{ mF}$ $\Delta h = 10\% (1 \text{ nom}, I_1 = 5 \dots 100\% I_1 \text{ nom}, Coxt = 0 \text{ mF}$ 0.350.6 VDC dV_1 ItLoad Transient Response dI_1 of $I \text{ nom}, I_1 = 5 \dots 10\% I_1 \text{ nom}, Coxt = 0 \text{ mF}$ 0.350.6 VDC dV_1 ItDynamic Load Regulation $\Delta h = 60\% I_1 \text{ nom}, I_1 = 5 \dots 10\% I_1 \text{ nom}, I_2 = 5 \dots 10\% I_1 \text{ nom}, I_2 = 5 \dots 10\% I_1 \text{ nom}, I_2 = 0 \dots 10\% I_1 \text{ nom}, I_1 = 5 \dots 167 \text{ A}, I_2 = 0 \dots 10\% I_1 \text{ nom}, I_1 = 5 \dots 100\% I_1 \text{ nom}, I_2 = 0 \dots 10\% I_1 \text{ nom}, I_3 = 0.6$ VDC V_1 dynDynamic Load Regulation $\Delta I_1 = 50 \dots 5000 \text{ Hz}$ puty cycle = 10 \dots 90\%, I11.412.6 V V_1 riseOutput Voltage Rise Time $V_1 = 10 \dots 90\% V_1 \text{ nom}, Cext < 10 \text{ mF}$ 130ms <t< td=""><td></td><td></td><td>Vi min LL to Vi max HL, 0 to 100% /1 nom,</td><td></td><td></td><td>120</td><td>mVpp</td></t<>			Vi min LL to Vi max HL, 0 to 100% /1 nom,			120	mVpp
$dV_{1 temp}$ Thermal Drift $0.5 \cdot l_{1 nom}, T_A = 0 \dots 55^{\circ} C$ -0.4 $mV/^{\circ} C$ $dl_{1 share}$ Current SharingDifference between individual l_{t_1} $1 \dots 8$ power supplies in parallel -6 $+6$ ADC V_{ISHARE} Current Share Bus Voltage $V_{SHARE at} 167A$ 8 VDC V_{ISHARE} Current Share Bus Voltage l_{peak} 9.14 VDC $dV_{1 lt}$ Load Transient Response $\Delta h = 50\% l_{1 nom}, l_{1} = 5 \dots 100\% l_{1 nom}, cext = 0 mF$ 0.35 0.6 VDC $dV_{1 lt}$ Load Transient Response $\Delta h = 10\% l_{1 nom}, l_{1} = 0 \dots 10\% l_{1 nom}, cext = 0 mF$ 0.35 0.6 VDC $dV_{1 lt}$ Load Transient Response $\Delta h = 60\% l_{1 nom}, l_{1} = 5 \dots 100\% l_{1 nom}, cext = 0 mF$ 0.35 0.6 VDC $dV_{1 lt}$ Cext = 0 mF $\Delta h = 60\% l_{1 nom}, l_{1} = 5 \dots 10\% l_{1 nom}, cext = 10 \dots 90\%, 11.4$ 12.6 V $dV_{1 lt}$ Dynamic Load Regulation $f = 50 \dots 5000 Hz, Duty cycle = 10 \dots 90\%, 11.4$ 12.6 V $V_{1 dyn}$ Dynamic Load Regulation $V_{1 = 0} \dots 20\% V_{1 nom}, cext < 10 mF$ 1 30 ms $t_{V1 rise}$ Output Voltage Rise Time $V_{1 = 0} \dots 90\% V_{1 nom}, cext < 10 mF$ 1 30 ms $t_{V1 ovr sh}$ Output Turn-on Overshoot0 to 100\% l_{1 nom} 0.6 V $dV_{1 sense}$ Remote SenseCompensation for cable drop, 0 to 100\% l_{1 nom} 0.25 V	dV1 load	Load Regulation	0 to 100% <i>I</i> 1 nom	-83	-110	-138	mV
dl_1 shareCurrent SharingDifference between individual l_1 , 1 8 power supplies in parallel-6+6ADC V_{ISHARE} Current Share Bus Voltage V_{ISHARE} at 167A8VDC V_{ISHARE} Current Share Bus Voltage $l_1 peak$ 9.14VDC dV_1 ItLoad Transient Response $\Delta h = 50\% l_1 nom, l_1 = 5 100\% l_1 nom, compared to mF0.350.6VDCdV_1 It\Delta h = 10\% l_1 nom, l_1 = 0 10\% l_1 nom, cext = 0 mF0.350.6VDCdV_1 It\Delta h = 60\% l_1 nom, l_1 = 5 100\% l_1 nom, cext = 0 mF0.351msdV_1 It\Delta h = 60\% l_1 nom, l_1 = 5 107 A, l_1 = 5 107 A, l_1 = 5 107 A, l_2 = 3 107 A, l_2 = 3 107 A, l_2 = 3 100\% l_1 nom, cext < 10 mF$	dV1 line	Line Regulation	Vimin HL to Vimax HL, 0.5 · /1 nom	-24	0	24	mV
df shareCurrent Sharing1 8 power supplies in parallel-646ADC V_{ISHARE} Current Share Bus Voltage V_{ISHARE} at 167A8VDC V_{ISHARE} Current Share Bus Voltage I_{Ipeak} 9.14VDC $dV_{1 It}$ Load Transient Response $\Delta h = 50\% I_{1 nom}, I_{1} = 5 100\% I_{1 nom}, 0.35$ 0.6VDC $dV_{1 It}$ Load Transient Response $\Delta h = 10\% I_{1 nom}, I_{1} = 0 10\% I_{1 nom}, 0.35$ 0.6VDC $dV_{1 It}$ Dynamic Load Regulation $Ah = 60\% I_{1 nom}, I_{1} = 5 167 A, f = 50 5000 Hz, Duty cycle = 10 90\%, 11.4$ 12.6V $V_{I dyn}$ Dynamic Load Regulation $f = 50 5000 Hz, Duty cycle = 10 90\%, 11.4$ 12.6V $V_{I dyn}$ Output Voltage Rise Time $V_{I} = 1090\% V_{I nom}, Cext < 10 mF$ 130ms $t_{VI ovr sh}$ Output Turn-on Overshoot0 to 100% $I_{I nom}$ 0.6VV $dV_{I sense}$ Remote SenseCompensation for cable drop, 0 to 100% $I_{I nom}$ 0.25V	dV1 temp	Thermal Drift			-0.4		mV/°C
VISHARECurrent Share Bus Voltage $I_{1 peak}$ 9.14VDC $dV_{1 lt}$ Load Transient Response $\Delta h = 50\% I_{1 nom,} I_{1} = 5 \dots 100\% I_{1 nom,}$ 0.350.6VDC $dV_{1 lt}$ Load Transient Response $\Delta h = 10\% I_{1 nom,} I_{1} = 0 \dots 10\% I_{1 nom,}$ 0.350.6VDC $dV_{1 lt}$ $Cext = 0 mF$ $\Delta h = 10\% I_{1 nom,} I_{1} = 0 \dots 10\% I_{1 nom,}$ 0.350.6VDC t_{rec} Recovery Time $dI_{1} dt = 1A/\mu s,$ recovery within 1% of $V_{1 nom}$ 0.51ms $V_{1 dyn}$ Dynamic Load Regulation $f = 50 \dots 5000 Hz,$ Duty cycle = 10 \dots 90\%, 11.412.6V $t_{V1 dyn}$ Output Voltage Rise Time $V_{1} = 10 \dots 90\% V_{1 nom,} Cext < 10 mF$ 130ms $t_{V1 ovr sh}$ Output Turn-on Overshoot0 to 100% $I_{1 nom}$ 0.6V $dV_{1 sense}$ Remote SenseCompensation for cable drop, 0 to 100% $I_{1 nom}$ 0.25V	dl1 share	Current Sharing	,	-6		+6	ADC
$dV_{1 \ lt}$ $\Delta h = 50\% \ lt \ nom, \ lt = 5 \dots 100\% \ lt \ nom, \ cext = 0 \ mF$ 0.35 0.6 VDC $dV_{1 \ lt}$ $\Delta h = 10\% \ lt \ nom, \ lt = 0 \dots 10\% \ lt \ nom, \ cext = 0 \ mF$ 0.35 0.6 VDC $dV_{1 \ lt}$ $\Delta h = 10\% \ lt \ nom, \ lt = 0 \dots 10\% \ lt \ nom, \ cext = 0 \ mF$ 0.35 0.6 VDC t_{rec} Recovery Time $dh/dt = 1A/\mu s, recovery within 1% \ of \ V_{1 \ nom}$ 0.5 1 ms $V_{1 \ dyn}$ Dynamic Load Regulation $f = 50 \dots 5000 \ lt \ nom, \ lt = 5 \dots 167 \ A, \ cext = 2 \dots 30 \ mF$ 11.4 12.6 V $t_{V1 \ dyn}$ Output Voltage Rise Time $V_1 = 10 \dots 90\% \ V_1 \ nom, \ Cext < 10 \ mF$ 1 30 ms $t_{V1 \ our \ sh}$ Output Turn-on Overshoot $0 \ to 100\% \ lt \ nom$ 0.6 V $dV_1 \ sense$ Remote SenseCompensation for cable drop, $0 \ to 100\% \ lt \ nom$ 0.25 V	VISHARE	Current Share Bus Voltage	V _{ISHARE at} 167A		8		VDC
$dV_{1:tr}$ $C_{ext} = 0 \text{ mF}$ 0.35 0.6 VDC $dV_{1:tr}$ $\Delta h = 10\% h_{1:nom}, h_{1} = 0 \dots 10\% h_{1:nom}, C_{ext} = 0 \text{ mF}$ 0.35 0.6 VDC t_{rec} Recovery Time $dh/dt = 1A/\mu s, recovery within 1\% of V_{1:nom}0.350.6VDCt_{rec}Recovery Timedh/dt = 1A/\mu s, recovery within 1\% of V_{1:nom}0.51msV_{1:dyn}Dynamic Load Regulationf = 50 \dots 5000 \text{ Hz}, \text{Duty cycle} = 10 \dots 90\%, 11.412.6Vt_{V1:rise}Output Voltage Rise TimeV_{1:nom}, C_{ext} < 10 \text{ mF}130mst_{V1:rise}Output Voltage Rise TimeV_{1:nom}, C_{ext} < 10 \text{ mF}130mst_{V1:rise}Output Voltage Rise TimeV_{1:nom}0.6VdV_{1:sense}Remote SenseCompensation for cable drop, 0 to 100% h_{1:nom}0.25V$	VISHARE	Current Share Bus Voltage	It peak		9.14		VDC
Load Transient Response $\Delta h = 10\% I_{1 nom}, I_{1} = 0 \dots 10\% I_{1 nom}, C_{ext} = 0 \text{ mF}$ 0.350.6VDC $dV_{1 tt}$ $C_{ext} = 0 \text{ mF}$ 0.350.6VDC t_{rec} Recovery Time $dh/dt = 1A/\mu \text{s}, \text{ recovery within } 1\% \text{ of } V_{1 nom}$ 0.51ms $V_{1 dyn}$ Dynamic Load Regulation $\Delta h = 60\% I_{1 nom}, I_{1} = 5 \dots 167 \text{ A}, I_{2} = 50 \dots 5000 \text{ Hz}, \text{ Duty cycle } = 10 \dots 90\%, I_{1.4}$ 11.412.6V $t_{V1 dyn}$ Output Voltage Rise Time $V_{1} = 10 \dots 90\% V_{1 nom}, C_{ext} < 10 \text{ mF}$ 130ms $t_{V1 our sh}$ Output Turn-on Overshoot0 to 100\% I_{1 nom}0.6V $dV_{1 sense}$ Remote SenseCompensation for cable drop, 0 to 100% I_{1 nom}0.25V	dV1 It				0.35	0.6	VDC
$V_{I \ dyn}$ Dynamic Load Regulation $\Delta h = 60\% \ l_{I \ nom}, l_{I} = 5 \dots 167 \ A,$ $f = 50 \dots 5000 \ Hz, Duty cycle = 10 \dots 90\%,$ $C_{ext} = 2 \dots 30 \ mF$ 11.412.6V $t_{VI \ nise}$ Output Voltage Rise Time $V_{I} = 10 \dots 90\% \ V_{I \ nom}, C_{ext} < 10 \ mF$ 130ms $t_{VI \ our \ sh}$ Output Turn-on Overshoot0 to 100\% \ l_{I \ nom}0.6V $dV_{I \ sense}$ Remote SenseCompensation for cable drop, 0 to 100% \ l_{I \ nom}0.25V	dV1 It	Load Transient Response	$\Delta h = 10\% I_{1 \text{ nom}}, I_{1} = 0 \dots 10\% I_{1 \text{ nom}},$		0.35	0.6	VDC
$V_{1 dyn}$ Dynamic Load Regulation $f = 50 \dots 5000$ Hz, Duty cycle = $10 \dots 90\%$, 11.4 12.6 V $t_{V1 rise}$ Output Voltage Rise Time $V_1 = 10 \dots 90\%$ $V_{1 nom}$, $C_{ext} < 10$ mF 1 30 ms $t_{V1 ovr sh}$ Output Turn-on Overshoot 0 to 100% $I_{1 nom}$ 0.6 V dV_1 senseRemote SenseCompensation for cable drop, 0 to 100% $I_{1 nom}$ 0.25 V	trec	Recovery Time	$dh/dt = 1A/\mu s$, recovery within 1% of $V_{1 nom}$		0.5	1	ms
tv1 ovr shOutput Turn-on Overshoot0 to 100% /1 nom0.6VdV1 senseRemote SenseCompensation for cable drop, 0 to 100% /1 nom0.25V	V1 dyn	Dynamic Load Regulation	<i>f</i> = 50 5000 Hz, Duty cycle = 10 90%,	11.4		12.6	V
<i>dV_{1 sense}</i> Remote Sense Compensation for cable drop, 0 to 100% <i>I_{1 nom}</i> 0.25 V	tv1 rise	Output Voltage Rise Time	V ₁ = 1090% V _{1 nom} , C _{ext} < 10 mF	1		30	ms
	t _{V1 ovrsh}	Output Turn-on Overshoot	0 to 100% <i>I</i> _{1 nom}			0.6	V
C _{V1 load} Capacitive Loading 0 30 mF	dV1 sense	Remote Sense	Compensation for cable drop, 0 to 100% I1 nom			0.25	V
	$C_{V1 \ load}$	Capacitive Loading		0		30	mF

⁸ Measured with a 10 uF low ESR capacitor in parallel with a 0.1 uF ceramic capacitor at the point of measurement



⁶ See also chapter <u>TEMPERATURE AND FAN CONTROL</u>

⁷ Peak combined power for all outputs must not exceed 2100 W; maximum of peak power duration is 20 seconds without asserting the SMBAlert signal

5.2 STANBY OUTPUT VSB

General Condition: $T_A = 0...40$ °C (PET2000-12-074RA), $T_A = 0...55$ °C (PET2000-12-074NA), Vi = 230 VAC unless otherwise noted.

PARAMET	ER	DESCRIPTION / CONDIT	ION	MIN	NOM	MAX	UNIT
VSB nom	Nominal Output Voltage	<i>IsB = 0 A</i> . <i>T</i> ₄ = 25°C			12.1		VDC
V _{SB set}	Output Setpoint Accuracy	$T_{SB} = 0$ A, $T_A = 25$ O		-1		+1	%V _{SBnom}
dV _{SB tot}	Total Regulation	$V_{i \min LL}$ to $V_{i \max HL}$, 0 to 100%	6 I _{SB nom}	-5		+1	%V _{SBnom}
P _{SB nom}	Nominal Output Power	Vimin LL to Vimax HL	PET2000-12-074RA PET2000-12-074NA		36 60		W W
P _{SB peak}	Peak Output Power ⁸	$V_{i \min LL}$ to $V_{i \max HL}$	PET2000-12-074RA PET2000-12-074NA		36 60		W
ISB nom	Output Current	Vimin LL to Vimax HL	PET2000-12-074RA PET2000-12-074NA	0		3 5	ADC
ISB peak	Peak Output Current9	Vimin LL to Vimax HL	PET2000-12-074RA PET2000-12-074NA	0		3.3 5.3	ADC
V _{SB pp}	Output Ripple Voltage ⁷	<i>Vi min LL</i> to <i>Vi max HL</i> , 0 to 100%	$6 I_{SB nom}, C_{ext} = 0 \text{ mF}$			120	mVpp
dVsB load	Load Regulation	0 to 100% IsB nom	PET2000-12-074RA PET2000-12-074NA	-144 -290	-240 -430	-330 -570	mV mV
dV _{SB line}	Line Regulation	$V_{i \min HL}$ to $V_{i \max HL}$, $I_{SB nom} = 0$	0 A	-24	0	24	mV
dVsB temp	Thermal Drift	$I_{SB} = 0 A$			-0.5		mV/°C
dlsB share	Current Sharing	Deviation from ISB tot / N, ISB	= 0.5 · <i>IsB nom</i>	-1		+1	ADC
VsB dyn	Load Transient Response	Δ <i>IsB</i> = 50% <i>IsB nom</i> , <i>IsB</i> = 5	. 100% <i>IsB nom</i> ,		0.2	0.3	VDC
trec	Recovery Time	$dI_{SB}/dt = 1A/\mu s$, recovery wi	thin 1% of V _{SB nom}		1	2	ms
VsB dyn	Dynamic Load Regulation	$\Delta I_{SB} = 1 \text{A}, I_{SB} = 0 \dots I_{SB \text{ nom}},$ Duty cycle = 10 \dots 90%, Cex		11.4		12.6	V
t _{VSB rise}	Output Voltage Rise Time	$V_{SB} = 1090\% V_{SB nom}, C_{ext}$	< 1 mF	1	2	5	ms
tvsB ovr sh	Output Turn-on Overshoot	0 to 100% <i>IsB nom</i>				0.6	V
$C_{VSB \ load}$	Capacitive Loading			0		3100	μF

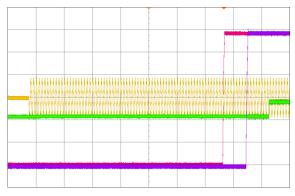


Figure 5. Turn-On AC Line 230VAC, full load (200ms/div) CH1: Vin (400V/div) CH2: PWOK_H (5V/div) CH3: V₁ (2V/div) CH4: V_{SB} (2V/div)

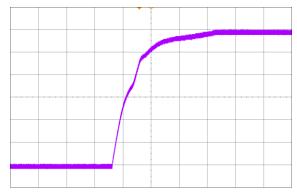


Figure 6. Rise time V1 at 230VAC, full load (2ms/div) CH3: V1 (2V/div)

⁹ In single power supply configuration



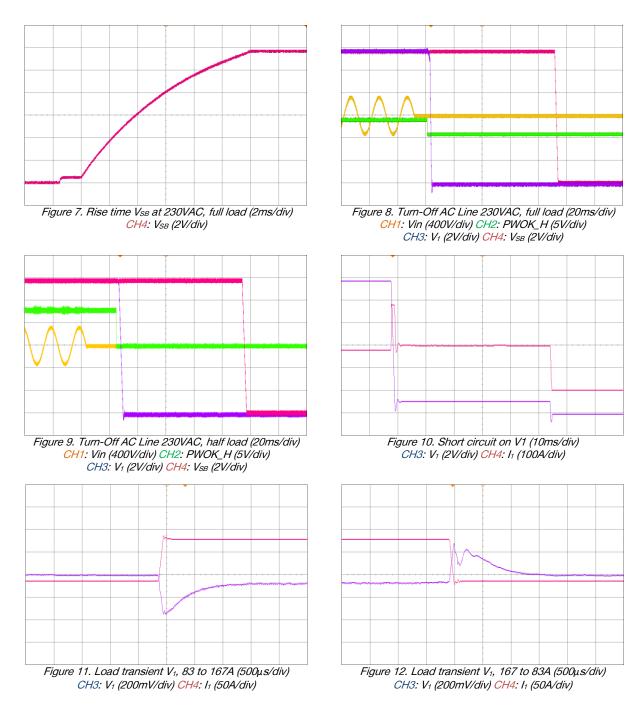
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5.3 OUTPUT GROUND / CHASSIS CONNECTION

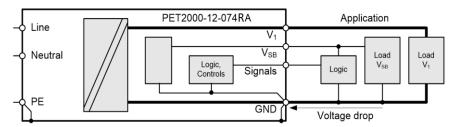
The output return path serves as power and signal ground. All output voltages and signals are referenced to these pins. To prevent a shift in signal and voltage levels due to ground wiring voltage drop a low impedance ground plane should be used as shown in *Figure 13.* Alternatively, separated ground signals can be used as shown in

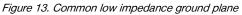
Figure 14. In this case the two ground planes should be connected together at the power supplies ground plane.



NOTE:

Within the power supply the output GND pins are connected to the Chassis, which in turn is connected to the Protective Earth terminal on the AC inlet. Therefore, it is not possible to set the potential of the output return (GND) to any other than Protective Earth potential.





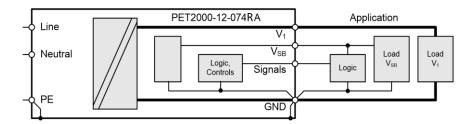


Figure 14. Separated power and signal ground

6. PROTECTION

PARAME	ſER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
F	Input fuse (L)	Not use accessible, time-lag (T)		16		А
V1 OV	OV Threshold V1	Quar Valtage 1/ Drotaction Latab off Tune	13.3	13.9	14.5	VDC
t _{V1 OV}	OV Trip Time V_1	Over Voltage V ₇ Protection, Latch-off Type			1	ms
Vsb ov	OV Threshold VSB	Quer Veltara I/ Protection Automatic rate acab 1	13.3	13.9	14.5	VDC
tvsø ov	OV Trip Time VSB	Over Voltage V ¹ Protection, Automatic retry each 1s			1	ms
,	OC Limit V1	Over Current Limitation, Latch-off, Vimin HL to Vimax HL	169		175	ADC
I1 OC Slow		Over Current Limitation, Latch-off, Vimin LL to Vimax LL	85		88	ADC
tv1 OC Slow	OC Trip time V_1	Over Current Limitation, Latch-off time	20			s
,	Fact 00 Limit 1/	Fast Over Current Limit., Latch-off, $V_{i \min HL}$ to $V_{i \max HL}$	176		180	ADC
I _{V1 OC Fast}	Fast OC Limit V_1	Fast Over Current Limit., Latch-off, Vimin LL to Vimax LL	110		115	ADC
tv1 OC Fast	Fast OC Trip time V1	Fast Over Current Limitation, Latch-off time	50	55	60	ms
lı sc	Max Short Circuit Current 1/1	<i>V</i> ₁ < 3 V			180	А
tv1 sc	Short Circuit Regulation Time	$V_1 < 3$ V, time until I_1 is limited to $< I_{1 sc}$			2	ms
Isb oc	OC Limit VsB	Over Current Limitation, Constant-Current Type	5.2		7.5	А
tvsø oc	OC Trip time VSB	Over Current Limit., time until ISB is limited to ISB OC			1	ms
T _{SD}	Over Temperature	See chapter 10.2				°C



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6.1 OVERVOLTAGE PROTECTION

PET2000-12-074xA front-end provides a fixed threshold overvoltage (OV) protection implemented with a HW comparator for both the main and the standby output. Once an OV condition has been triggered on the main output, the supply will shut down and latch the fault condition. The latch can be unlocked by disconnecting the supply from the AC mains or by toggling the PSON_L input. The standby output will continuously try to restart with a 1 s interval after OV condition has occurred.

6.2 UNDERVOLTAGE DETECTION

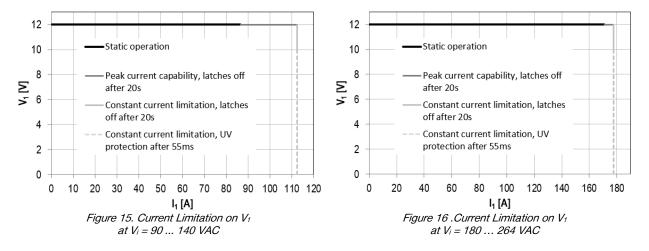
Both main and standby outputs are monitored. LED and PWOK_H pin signal if the output voltage exceeds ±5% of its nominal voltage.

The main output will latch off if the main output voltage V_7 falls below 10 V (typically in an overload condition) for more than 55 ms. The latch can be unlocked by disconnecting the supply from the AC mains or by toggling the PSON_L input. If the standby output leaves its regulation bandwidth for more than 2 ms then the main output is disabled to protect the system.

6.3 CURRENT LIMITATION

MAIN OUTPUT

The main output exhibits a substantially rectangular output characteristic controlled by a software feedback loop. If output current exceeds $I_{V1 OC Fast}$ it will reduce output voltage in order to keep output current at $I_{V1 OC Fast}$. If the output voltage drops below ~10.0 VDC for more than 55 ms, the output will latch off (standby remains on), see also *Undervoltage Detection*.



A second SW controlled current limit will latch off the main output if the power supply is operated for long duration in its peak current capability region. This protection trips as soon as the output current exceeds $I_{1 OC Slow}$ for a duration of more than 20 s. The third current limitation implemented as a fast hardware circuit will immediately switch off the main output if the output current increases beyond the peak current trip point, occurring mainly if a short circuit is applied to the output voltage. The supply will re-start 4 ms later with a soft start, if the short circuit persists ($V_1 < 10.0$ V for >55 ms) the output will latch off; otherwise it continuous to operate.

The latch can be unlocked by disconnecting the supply from the AC mains or by toggling the PSON_L input.

The main output current limitation thresholds for $I_{1 OC Slow}$ and $I_{1 OC Fast}$ depend on the actual input voltage range applied to the power supply. In addition, the threshold for $I_{1 OC Slow}$ is reduced when ambient temperature exceeds 55°C, see *Figure 38* for PET2000-12-074RA and *Figure 46* for PET2000-12-074NA.

STANDBY OUTPUT

The standby output exhibits a substantially rectangular output characteristic down to 0 V (no hiccup mode / latch off). The current limitation of the standby output is independent of the AC input voltage.

Running in current limitation causes the output voltage to fall, this will trigger under voltage protection and disables the main output, see also Undervoltage Detection.



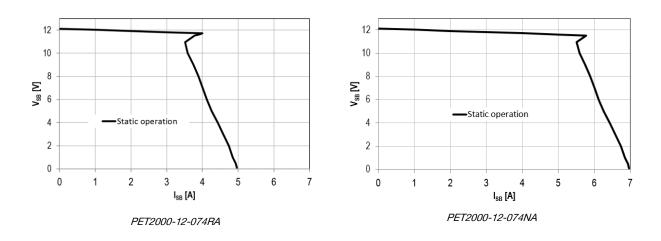


Figure 17. Current Limitation on V_{SB}

7. MONITORING

The power supply operating parameters can be accessed through I²C interface. For more details refer to chapter I2C / POWER MANAGEMENT BUS COMMUNICATION and document URP.00234 (PET Front-End Power Management Bus Communication Manual).

PARAME	TER	DESCRIPTION / CONDITION	MIN NOM	MAX	UNIT
V _{i mon}	Input RMS Voltage	$V_{i \min LL} \leq V_i \leq V_{i \max HL}$	-3	+3	VAC
,	Input DMC Current	<i>li</i> > 6.7 Arms	-3	+3	%
li mon	Input RMS Current	<i>l</i> _i ≤ 6.7 Arms	-0.2	+0.2	Arms
Pimon True Input Power	<i>P</i> _i > 500 W	-4	+4	%	
	50 W < $P_i \le$ 500 W	-20	+20	W	
V1 mon	V1 Voltage		-0.1	+0.1	VDC
,	V/4 Ourmant	/1 > 50 A	-1	+1	%
I _{1 mon}	V1 Current	$5 \text{ A} < I_7 \le 50 \text{ A}$	-0.5	+0.5	ADC
D	V/1 Output Dower	<i>P</i> _i > 1000 W	-1	+1	%
P _{1 nom}	V1 Output Power	50 W < P_i ≤ 1000 W	-10	+10	W
VSB mon	VSB Voltage		-0.1	+0.1	VDC
ISB mon	VSB Current		-0.1	+0.1	ADC
T _{A mon}	Inlet Temperature	$T_{A \min} \leq T_{A} \leq T_{A \max}$	-2	+2	°C



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8. SIGNALING AND CONTROL

8.1 ELECTRICAL CHARACTERISTICS

PARAME	TER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
PSON_H /	HOTSTANDBYEN_H					
VıL	Input Low Level Voltage	PSON_L: Main output enabled HOTSTANDBYEN_H: Hot Standby mode not allowed	-0.2		0.8	V
Vıн	Input High Level Voltage	PSON_L: Main output disabled HOTSTANDBYEN_H: Hot Standby mode allowed	2		3.5	V
I _{IL,H}	Maximum Input Sink or Source Current	<i>V</i> ₁ = -0.2 V to +3.5 V	-1		1	mA
R _{pull up}	Internal Pull up Resistor to internal 3.3 V			10		kΩ
RLOW	Maximum external Pull down Resistance to GND to obtain Low Level				1	kΩ
Rhigh	Minimum external Pull down Resistance to GND to obtain High Level		50			kΩ
PWOK_H						
Vol	Output Low Level Voltage	V_1 or V_{SB} out of regulation, $V_{Isink} < 4 \text{ mA}$	0		0.4	V
Vон	Output High Level Voltage	V_{1} and V_{SB} in regulation, $I_{source} < 0.5$ mA	2.4		3.5	V
Rpull up	Internal Pull up Resistor to internal 3.3 V			1		kΩ
IOL	Maximum Sink Current	<i>V</i> ₀ < 0.4 V			4	mA

8.2 SENSE INPUTS

The main output has sense lines implemented to compensate for voltage drop on load wires in both positive and negative path. The maximum allowed voltage drop is 200 mV on the positive rail and 50 mV on the GND rail.

With open sense inputs the main output voltage will rise by 270 mV. Therefore, if not used, these inputs should be connected to the power output and GND at the power supply connector. The sense inputs are protected against short circuit. In this case the power supply will shut down.

8.3 CURRENT SHARE

The PET front-ends have an active current share scheme implemented for V_{\prime} . All the ISHARE current share pins need to be interconnected in order to activate the sharing function. If a supply has an internal fault or is not turned on, it will disconnect its ISHARE pin from the share bus. This will prevent dragging the output down (or up) in such cases.

The current share function uses an analog bus to transmit and receive current share information. The controller implements a Master/Slave current share function. The power supply providing the largest current among the group is automatically the Master. The other supplies will operate as Slaves and increase their output current to a value close to the Master by slightly increasing their output voltage. The voltage increase is limited to +250 mV.

The standby output uses a passive current share method (droop output voltage characteristic).

8.4 PSON_L INPUT

The PSON_L is an internally pulled-up (3.3 V) input signal to enable/disable the main output V_7 of the front-end. With low level input the main output is enabled. This active-low pin is also used to clear any latched fault condition. The PSON_L can be either controlled by an open collector device or by a voltage source.



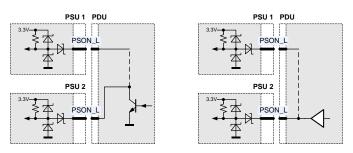


Figure 18. PSON_L connection

8.5 PWOK_H OUTPUT

The PWOK_H is an open drain output with an internal pull-up to 3.3 V indicating whether both V_{SB} and V_{7} outputs are within regulation. This pin is active-high.

An external pull down resistor ensures low level when there is no power supply seated. When combining PWOK_H outputs of several power supplies, circuits as shown in *Figure 19* should be used.

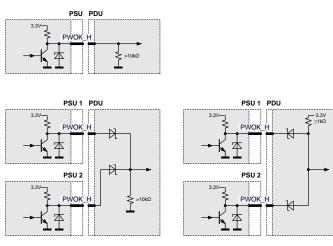


Figure 19. PWOK_H connection

8.6 HOT-STANDBY IN-/OUTPUT

The hot-standby operation is an operating mode allowing to further increase efficiency at light load conditions in a redundant power supply system. Under specific conditions one of the power supplies is allowed to disable its DC/DC stage. This will save the power losses associated with this power supply and at the same time the other power supply will operate in a load range having a better efficiency. In order to enable the hot standby operation, the HOTSTANDBYEN_H and the ISHARE pins need to be interconnected between the power supplies. A power supply will only be allowed to enter the hot-standby mode, when the HOTSTANDBYEN_H pin is high, the load current is low, see *Figure 20*, and the supply was allowed to enter the hot-standby mode by the system controller via the appropriate I²C command (by default disabled). The system controller needs to ensure that only one of the power supplies is allowed to enter the hot-standby mode.

If a power supply is in a fault condition, it will pull low its active-high HOTSTANDBYEN_H pin which indicates to the other power supply that it is not allowed to enter the hot-standby mode or that it needs to return to normal operation should it already have been in the hot-standby mode.

NOTE:

The system controller needs to ensure that only one of the power supplies is allowed to enter the hot-standby mode.

Figure 21 shows the achievable power loss savings when using the hot-standby mode operation. A total power loss reduction of approx. 10 W is achievable.

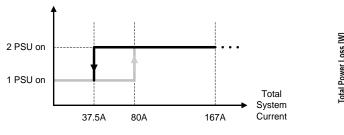


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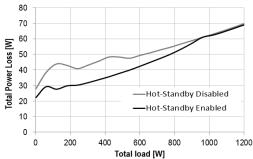


Figure 20. Hot-standby enable/disable current thresholds

Figure 21. PSU power losses with/without hot-standby mode

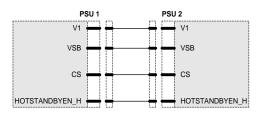


Figure 22. Recommended hot-standby configuration

8.7 PRESENT_L OUTPUT

8.8 SIGNAL TIMING

The PRESENT_L pin is wired through a 100 Ohms resistor to internal GND within the power supply. This pin does indicate that there is a power supply present in this system slot. An external pull-up resistor has to be added within the application. Current into PRESENT_L should not exceed 5 mA to guarantee a low level voltage if power supply is seated.

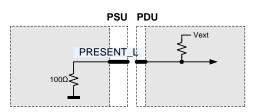


Figure 23. PRESENT_L connection

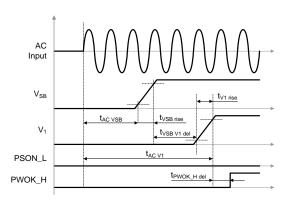


Figure 24. AC turn-on timing

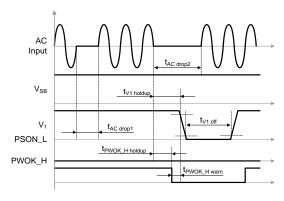


Figure 25. AC short dips



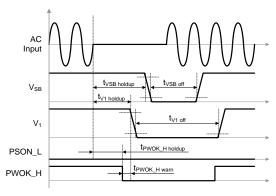


Figure 26. AC long dips

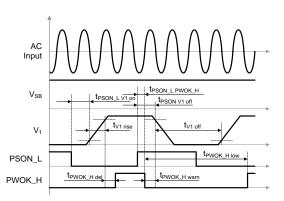


Figure 27. PSON_L turn-on/off timing

PARAMETER		DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
tac vsb	AC Line to 90% VsB				1.5	s
tac vi	AC Line to 90% V1	PSON_L = Low		1.5	4 ¹⁰	s
tvsB v1 del	V_{SB} to V_1 delay	PSON_L = Low	50	150	1000	ms
t _{V1 rise}	V ₁ rise time	See chapter OUTPUT				
t _{VSB rise}	V _{SB} rise time	See chapter OUTPUT				
		0.5 · I1 nom, ISB nom			17	ms
tAC drop1	AC drop without V_7 leaving regulation	0.7 · I1 nom, ISB nom			13	ms
		It nom, ISB nom			5	ms
tAC drop2	AC drop without VSB leaving regulation	It nom, ISB nom			70	ms
tv1 holdup	Loss of AC to V_7 leaving regulation	See chapter INPUT				
tvsB holdup	Loss of AC to VSB leaving regulation	See chapter INPUT				
tpwok_H del	Outputs in regulation to PWOK_H asserted		100	150	200	ms
tpwok_H warn	Warning time from de-assertion of PWOK_H to V_7 leaving regulation		0.15			ms
tpwok_H holdup	Loss of AC to PWOK_H de-asserted	Vi nom HL, I1 nom, ISB nom	10			ms
tpwok_H low	Time PWOK_H is kept low after being de-asserted		100			ms
tPSON_L V1 on	Delay PSON_L active to V_7 in regulation	$C_{ext} = 0 \text{ mF}$	5	10	20	ms
tpson_L V1 off	Delay PSON_L de-asserted to V_7 disabled		2	3	4	ms
tpson_l pwok_h	Delay PSON_L de-asserted to PWOK_H de-asserted			1	2	ms
t _{V1 off}	Time V_{7} is kept off after leaving regulation			1		S
tvsB off	Time VSB is kept off after leaving regulation			1		s

 $^{\rm 10}$ At repeated ON-OFF cycles the start-up times may increase by 1s



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8.9 LED INDICATOR

The front-end has one front LED showing the status of the supply. The LED is bi-colored: green and amber, and indicates AC and DC power presence and warning or fault conditions. *Table 1* lists the different LED status.

OPERATING CONDITION ¹¹	LED SIGNALING	
No AC or AC Line in UV condition, V_{SB} not present from paralleled power supplies	Off	
PSON_L High	Plinking Groep 1 Hz	
Hot-Standby Mode	Blinking Green 1 Hz	
No AC or AC Line in UV condition, V_{SB} present from paralleled power supplies		
V_1 or V_{SB} out of regulation		
Over temperature shutdown	Solid Amber	
Output over voltage shutdown (V_1 or V_{SB})		
Output over current shutdown (V_1 or V_{SB})		
Fan error (>15%)		
Over temperature warning	Blinking Amber 1 Hz	
Minor fan regulation error (>5%, <15%)		
Firmware boot loading in process	Blinking Green 2 Hz	
Outputs V_1 and V_{SB} in regulation	Solid Green	

Table 1. LED Status

9. I²C / POWER MANAGEMENT BUS COMMUNICATION

The PET front-end is a communication Slave device only; it never initiates messages on the I²C / SMBus by itself. The communication bus voltage and timing is defined in *Table 2* and further characterized through:

- The SDA/SCL IOs use 3.3 V logic levels
- External pull-up resistors on SDA/SCL required for correct signal edges
- Full SMBus clock speed of 100 kbps
- Clock stretching limited to 1 ms
- SCL low time-out of >25 ms with recovery within 10 ms
- Recognizes any time Start/Stop bus conditions

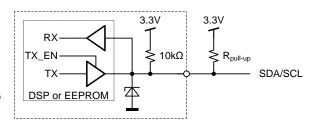


Figure 28. Physical layer of communication interface

Communication to the DSP or the EEPROM will be possible as long as the input AC voltage is provided. If no AC is present, communication to the unit is possible as long as it is connected to a life V_{SB} output (provided e.g. by the redundant unit). If only V_{7} is provided, communication is not possible.

¹¹ The order of the criteria in the table corresponds to the testing precedence in the controller



PARAMET	ER DESCRIPTION	CONDITION	MIN	MAX	UNIT
SCL / SDA					
ViL	Input low voltage		-0.5	1.0	V
Ин	Input high voltage		2.3	3.5	V
V _{hys}	Input hysteresis		0.15		V
V ₀L	Output low voltage	3 mA sink current	0	0.4	V
<i>t</i> r	Rise time for SDA and SCL		20+0.1Cb1	300	ns
<i>t</i> of	Output fall time ViHmin \rightarrow ViLmax	$10 \text{ pF} < C_{b}{}^{1} < 400 \text{ pF}$	20+0.1Cb1	250	ns
h	Input current SCL/SDA	0.1 VDD < Vi < 0.9 VDD	-10	10	μΑ
G	Internal Capacitance for each SCL/SDA			50	pF
<i>f</i> scl	SCL clock frequency		0	100	kHz
<i>R</i> pull-up	External pull-up resistor	f _{SCL} ≤ 100 kHz		1000 ns / C _b 1	Ω
<i>t</i> HDSTA	Hold time (repeated) START	f _{SCL} ≤ 100 kHz	4.0		μs
<i>t</i> Low	Low period of the SCL clock	f _{SCL} ≤ 100 kHz	4.7		μs
<i>t</i> HIGH	High period of the SCL clock	f _{SCL} ≤ 100 kHz	4.0		μs
t _{SUSTA}	Setup time for a repeated START	f _{SCL} ≤ 100 kHz	4.7		μs
<i>t</i> HDDAT	Data hold time	f _{SCL} ≤ 100 kHz	0	3.45	μs
<i>t</i> sudat	Data setup time	f _{SCL} ≤ 100 kHz	250		ns
<i>t</i> susto	Setup time for STOP condition	f _{SCL} ≤ 100 kHz	4.0		μs
<i>t</i> BUF	Bus free time between STOP and START	f _{SCL} ≤ 100 kHz	5		ms

¹ Cb = Capacitance of bus line in pF, typically in the range of 10...400 pF

Table 2. PC / SMBus Specification

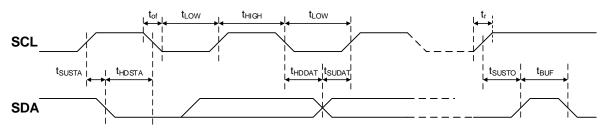


Figure 29. PC / SMBus Timing

ADDRESS SELECTION

The address for I²C communication can be configured by pulling address input pins A2, A1 and A0 either to GND (Logic Low) or leave them open (Logic High). An internal pull up resistor will cause the A2 / A1 / A0 pin to be in High Level if left open. A fixed addressing offset exists between the Controller and the EEPROM.



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A2	A1	A0	I2C Add	lress ¹²
AZ	AI	AU	Controller	EEPROM
0	0	0	0xB0	0xA0
0	0	1	0xB2	0xA2
0	1	0	0xB4	0xA4
0	1	1	0xB6	0xA6
1	0	0	0xB8	0xA8
1	0	1	0xBA	0xAA
1	1	0	0xBC	0xAC
1	1	1	0xBE	0xAE

Table 3. Address and protocol encoding

9.1 SMBALERT_L OUTPUT

The SMBALERT_L signal indicates that the power supply is experiencing a problem that the system agent should investigate. This is a logical OR of the Shutdown and Warning events. It is asserted (pulled Low) at Shutdown or Warning events such as reaching temperature warning/shutdown threshold of critical component, general failure, over-current, over-voltage, under-voltage or low-speed of a failed fan. This signal may also indicate the power supply is operating in an environment exceeding the specified limits.

The SMBAlert signal is asserted simultaneously with the LED turning to solid amber or blinking amber.

PARAM	ETER	R DESCRIPTION / CONDITION		NOM	MAX	UNIT
SMB_AL	LERT_L					
Vext	Maximum External Pull up Voltage				12	V
Іон	Maximum High Level Leakage Current	No Failure or Warning condition, V_{O} = 12 V			10	μA
Vol	Output Low Level Voltage	Failure or Warning condition, <i>Isink</i> < 4 mA	0		0.4	V
Rpull up	Internal Pull up Resistor to internal 3.3 V			None		
IOL	Maximum Sink Current	$V_O < 0.4 \text{ V}$			4	mA

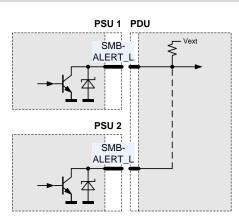


Figure 30. SMBALERT_L connection

¹² The LSB of the address byte is the R/W bit





9.2 CONTROLLER AND EEPROM ACCESS

The controller and the EEPROM in the power supply share the same I²C bus physical layer (see *Figure 31*) and can be accessed under different addresses, see ADDRESS SELECTION.

The SDA/SCL lines are connected directly to the controller and EEPROM which are supplied by internal 3.3 V.

The EEPROM provides 256 bytes of user memory. None of the bytes are used for the operation of the power supply.

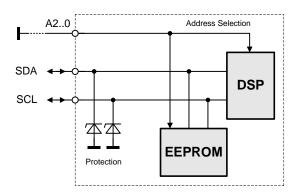


Figure 31. PC Bus to DSP and EEPROM

9.3 EEPROM PROTOCOL

The EEPROM follows the industry communication protocols used for this type of device. Even though page write / read commands are defined, it is recommended to use the single byte write / read commands.

WRITE

The write command follows the "SMBus 1.1 Write Byte Protocol". After the device address with the write bit cleared, the Two Byte Data Address is sent followed by the data byte and the STOP condition. A new START condition on the bus should only occur after 5ms of the last STOP condition to allow the EEPROM to write the data into its memory.

S Address W A	Data Address	Α	Data Address	Α	Data	AP	•
---------------	--------------	---	--------------	---	------	----	---

READ

The read command follows the "SMBus 1.1 Read Byte Protocol". After the device address with the write bit cleared the two byte data address is sent followed by a repeated start, the device address and the read bit set. The EEPROM will respond with the data byte at the specified location.



9.4 POWER MANAGEMENT BUS PROTOCOL

The Power Management Bus is an open standard protocol that defines means of communicating with power conversion and other devices. For more information, please see the System Management Interface Forum web site at: www.powerSIG.org.

Power Management Bus command codes are not register addresses. They describe a specific command to be executed. PET2000-12-074xA supply supports the following basic command structures:

- Clock stretching limited to 1 ms
- SCL low time-out of >25 ms with recovery within 10 ms
- Recognized any time Start/Stop bus conditions



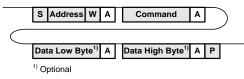
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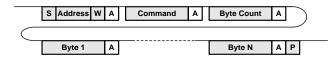
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WRITE

The write protocol is the SMBus 1.1 Write Byte/Word protocol. Note that the write protocol may end after the command byte or after the first data byte (Byte command) or then after sending 2 data bytes (Word command).

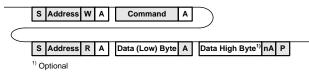


In addition, Block write commands are supported with a total maximum length of 255 bytes. See PET2000-12-074xA Power Management Bus Communication Manual URP.00234 for further information.

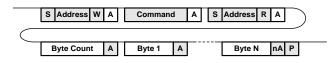


READ

The read protocol is the SMBus 1.1 Read Byte/Word protocol. Note that the read protocol may request a single byte or word.



In addition, Block read commands are supported with a total maximum length of 255 bytes. See PET2000-12-074xA Power Management Bus Communication Manual URP.00234 for further information.



9.5 GRAPHICAL USER INTERFACE

Bel Power Solutions provides with its "I²C Utility" a Windows® XP/Vista/Win7 compatible graphical user interface allowing the programming and monitoring of the PET2000-12-074xA Front-End.

The utility can be downloaded on: <u>belfuse.com/power-solutions</u> and supports both the PSMI and Power Management Bus protocols.

The GUI allows automatic discovery of the units connected to the communication bus and will show them in the navigation tree. In the monitoring view the power supply can be controlled and monitored.

If the GUI is used in conjunction with the YTM.00046 Evaluation Board it is also possible to control the PSON_L pin of the power supply. Refer to BCG.00809 for YTM.00046 connection and GUI configuration.



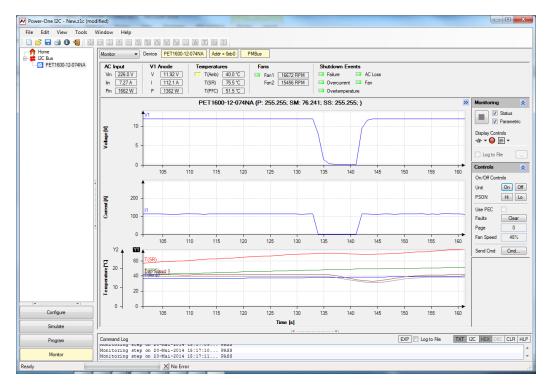
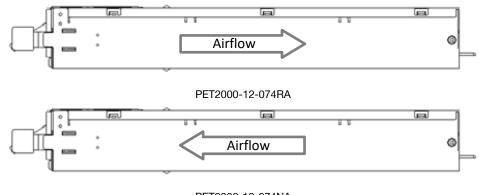


Figure 32. Monitoring dialog of the PC Utility

10. TEMPERATURE AND FAN CONTROL

10.1 FAN CONTROL

To achieve best cooling results sufficient airflow through the supply must be ensured. Do not block or obstruct the airflow at the rear of the supply by placing large objects directly at the output connector. The PET2000-12-074RA is provided with a front to rear airflow, which means the air enters on the AC-inlet side of the supply and leaves at the DC-output, while the PET2000-12-074NA is provided with a rear to front airflow, which means the air enters through the DC-output of the supply and leaves at the AC-inlet side, as shown in *Figure 33*. The PET2000-12-074xA supply has been designed for horizontal operation.



PET2000-12-074NA

Figure 33. Airflow direction

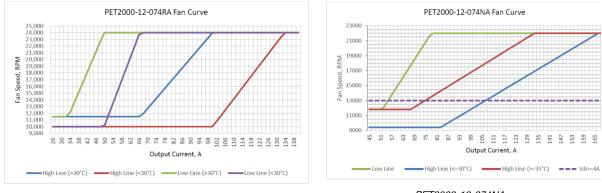


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PET2000-12-074xA

The fan inside the supply is controlled by a microprocessor. The rpm of the fan is adjusted to ensure optimal supply cooling and is a function of output power. Three different curves are selected based on input voltage and inlet temperature. With standby output loaded the fan speed minimum is limited to ensure enough cooling of circuits providing standby power. *Figure 34* illustrates the programmed fan curves.



PET2000-12-074RA

PET2000-12-074NA

Figure 34. Fan speed vs. main output load

10.2 TEMPERATURE MONITOR AND OVER TEMPERATURE PROTECTION

PET2000-12-074xA provides access via l^2C to the measured temperatures of in total 6 sensors within the power supply, see *Table 4*. The microprocessor is monitoring these temperatures and if warning threshold of one of these sensors is reached it will set fan to maximum speed. If temperatures continue to rise above shut down threshold the main output V_1 (or V_{SB} if auxiliary converter is affected) will be disabled. At the same time the warning or fault condition is signalized accordingly through LED, PWOK_H and SMBALERT_L.

TEMPERATURE SENSOR	DESCRIPTION / CONDITION	POWER MANAGEMENT BUS REGISTER	WARNING THRESHOLD	SHUTDOWN THRESHOLD
PET2000-12-074RA				
Inlet Air Temperature	Sensor located on control board close to DC end of PSU	8Dh	61°C	63°C
Synchronous Rectifier	Sensor located on secondary side of DC/DC stage	8Eh	105°C	110°C
Primary Heat Sink	Sensor located on primary heat sink	8Fh	96°C	101°C
Output ORing Element	Sensor located close to output	D2h	105°C	110°C
Auxiliary Converter	Sensor located on secondary side on auxiliary rectifier	D3h	95°C	100°C
Outlet Ambient	Sensor located near output connector	D4h	85°C	90°C
PET2000-12-074NA				
Inlet Air Temperature	Sensor located on control board close to DC end of PSU	8Dh	75°C	78°C
Synchronous Rectifier	Sensor located on secondary side of DC/DC stage	8Eh	95°C	100°C
Primary Heat Sink	Sensor located on primary heat sink	8Fh	87°C	92°C
Output ORing Element	Sensor located close to output	D2h	100°C	105°C
Auxiliary Converter	Sensor located on secondary side on auxiliary rectifier	D3h	80°C	85°C
Bridge Rectifier	Sensor located on heat sink for AC rectifier	D4h	86°C	91°C

Table 4. Temperature sensor location and thresholds



10.3 MAXIMUM OUTPUT POWER VERSUS INLET TEMPERATURE FOR SAFETY COMPLIANCY

For safety compliant operation the power supply must not exceed specified operating conditions specified herein. These operating conditions ensure the input AC connector is operated within its ratings.

The different input AC connectors and regional usage is not considered in this implementation of current limitation. Therefore, it is under the responsibility of the user to ensure safety compliant operation.

10.3.1 PET2000-12-074RA

Between 0°C and 40°C power supply inlet temperature the maximum allowed output power is only depending on AC input connector type chosen, regional usage and the applied nominal input AC voltage. Above 40°C the maximum output power is further reduced with rising temperature. *Figure 35* to *Figure 38* illustrate these maximum current and power levels.

The mentioned power levels are related to main output power only, in addition the standby output can be operated up to 5 A with derating to 3 A as shown in *Figure 37*.

Above 55°C the power supply is adjusting the current limit level $I_{1 OC Slow}$ depending on input voltage range (100-127 VAC or 200-240 VAC) and inlet temperature, as shown in *Figure 38* to protect the power supply from excessive component temperatures.

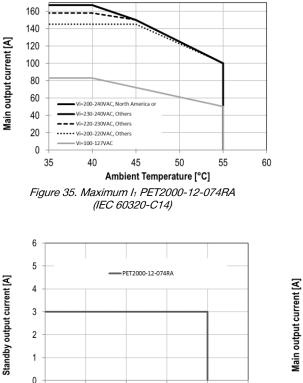


Figure 37. Maximum ISB

Ambient Temperature [°C]

45

50

55

60

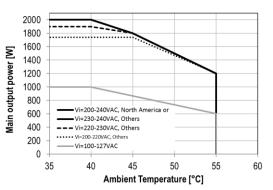


Figure 36. Maximum P1 PET2000-12-074RA (IEC 60320-C14)

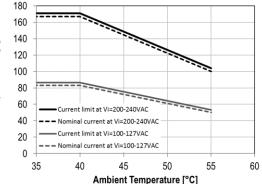


Figure 38. Current limitation vs temperature



35

40

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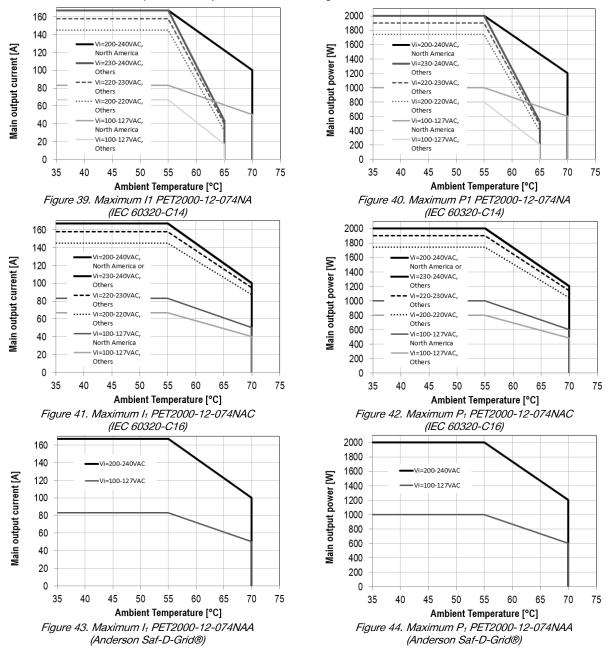
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10.3.2 PET2000-12-074NA

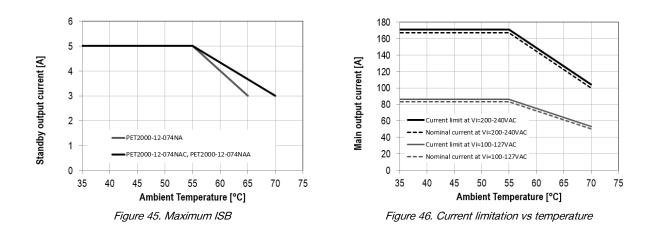
Between 0°C and 55°C power supply inlet temperature the maximum allowed output power is only depending on AC input connector type chosen, regional usage and the applied nominal input AC voltage. Above 55°C the maximum output power is further reduced with rising temperature. *Figure 39* to *Figure 44* illustrate these maximum current and power levels.

The mentioned power levels are related to main output power only, in addition the standby output can be operated up to 5 A with derating to 3 A as shown in *Figure 45*.

Above 55°C the power supply is adjusting the current limit level *I*_{1 OC Slow} depending on input voltage range (100-127 VAC or 200-240 VAC) and inlet temperature, as shown in *Figure 46*.







11. ELECTROMAGNETIC COMPATIBILITY

11.1 IMMUNITY

PARAMETER	DESCRIPTION / CONDITION	CRITERION
ESD Contact Discharge	IEC / EN 61000-4-2, ±8 kV, 25+25 discharges per test point (metallic case, LED, connector body)	А
ESD Air Discharge	IEC / EN 61000-4-2, ±15 kV, 25+25 discharges per test point (non-metallic user accessible surfaces)	А
Radiated Electromagnetics Filed	IEC / EN 61000-4-3, 10 V/m, 1 kHz/80% Amplitude Modulation, 1µs Pulse Modulation, 10 kHz 2 GHz	А
Burst	IEC / EN 61000-4-4, Level 3 AC port ±2 kV, 1 minute	А
Surge	IEC / EN 61000-4-5, Level 3 Line to Earth: ±2 kV Line to Line: ±1 kV	A
RF Conducted Immunity	IEC / EN 61000-4-6, Level 3, 10 Vrms, CW, 0.1 80 MHz	А
Voltage Dips and Interruptions	IEC / EN 61000-4-11 Vi 230VAC / 50 Hz, 90% load, Phase 0°, Dip 100% , duration 10 ms Vi 200VAC / 50 Hz, 70% load, Phase 0°, Dip 30% , duration 500 ms Vi 200VAC / 50 Hz, 100% load, Phase 0°, Dip 20% , duration 10 s	V1: A, V5B: A V1: A, V5B: A V1: A, V5B: A

11.2 EMISSION

PARAMETER	DESCRIPTION / CONDITION	CRITERION
Conducted Emission	EN 55022 / CISPR 22: 0.15 30 MHz, QP and AVG, single power supply EN 55022 / CISPR 22: 0.15 30 MHz, QP and AVG, 2 power supplies in a system	Class A 6 dB margin Class A
Radiated Emission	EN 55022 / CISPR 22: 30 MHz 1 GHz, QP, single power supply EN 55022 / CISPR 22: 30 MHz 1 GHz, QP, 2 power supplies in a system	Class A 6 dB margin Class A
Harmonic Emissions	IEC 61000-3-2, Vi = 115 VAC / 60 Hz & 230 VAC / 50 Hz, 100% Load	Class A
AC Flicker	IEC 61000-3-3, Vi = 230 VAC / 50Hz, 100% Load	Pass
Acoustical Noise	Distance at bystander position, 25°C, 50% Load	65 dBA



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12. SAFETY / APPROVALS

Maximum electric strength testing is performed in the factory according to IEC/EN 60950, and UL 60950. Input-to-output electric strength tests should not be repeated in the field. Bel Power Solutions will not honor any warranty claims resulting from electric strength field tests.

PARAMETER	DESCRIPTION / CONDITION	NOTES
Agency Approvals	Approved to latest edition of the following standards: UL/CSA60950-1, IEC60950-1 and EN60950-1. NEMKO NO86275, EAC NO 0230738, COC	Approved
Grade of Insulation	Input (L/N) to chassis (PE) Input (L/N) to output Output to chassis	Basic Reinforced None (Direct connection)
Creepage / Clearance	Primary (L/N) to chassis (PE) Primary to secondary	
Electrical Strength Test	Input to chassis Input to output (tested by manufacturer only)	Min. 2121 VDC 4242 VDC

13. ENVIRONMENTAL

PARA	METER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
TA	Ambient Temperature	Up to 1'000 m ASL	0		+40 +55*	°C
74		Linear derating from 1'000 to 3'048 m ASL			+35 +45*	°C
-	Enternal of Terror Denses	Reduced output power ¹³ , up to 1'000 m ASL			+55 +70*	°C
T _{A ext}	Extended Temp. Range	Linear derating from 1'000 to 3'048 m ASL			+50 +60*	
T_S	Storage Temperature	Non-operational	-20		+70	°C
	Altitude	Operational, above Sea Level	-		3'048	m
	Annude	Non-operational, above Sea Level	-		10'600	m
	Shock, operational	Half sine, 11ms, 10 shocks per direction,			1	g peak
	Shock, non-operational	6 directions			30	g peak
	Vibration, sinusoidal, operational	IEC/EN 60068-2-6, sweep 5 to 500 to 5 Hz,			1	g peak
	Vibration, sinusoidal, non-operational	1 octave/min, 5 sweep per axis			4	g peak
	Vibration, random, non-operational	IEC/EN 60068-2-64, 5 to 500 Hz, 1 hour per axis			0.025	g²/Hz

* Max temperature values for PET2000-12-074NA model.

14. RELIABILITY

PARAMETER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
<i>MTBF</i> Mean time to failure	$T_A = 25^{\circ}$ C, according Telcordia SR-332, issue 3, GB, confidence level = 90%	860			kh

15. MECHANICAL

PARA	METER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
		Width		73.5		mm
	Dimensions	Heigth		40.0		mm
		Depth		265.0		mm
m	Weight			1.1		kg

¹³ See chapter 10.3





15.1 OUTLINE PET2000-12-074xA, PET2000-12-074xAC

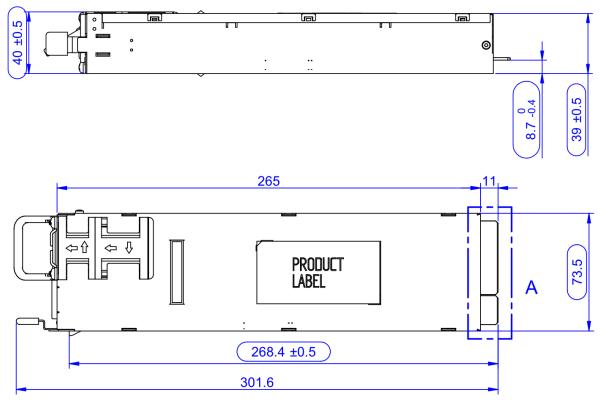


Figure 47. Top and side view

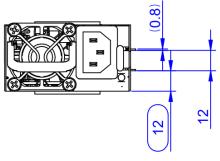


Figure 48. Front view

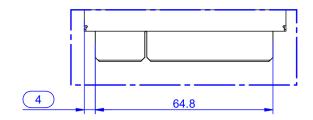


Figure 49. Rear view



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15.2 OUTLINE PET2000-12-074NAA

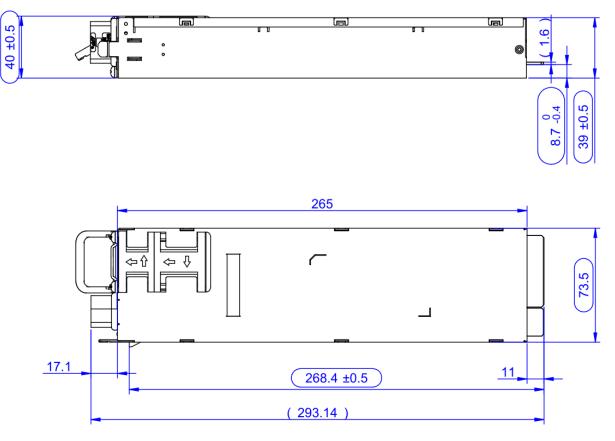


Figure 50. Top and side view

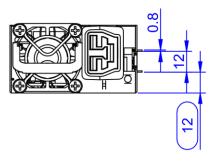
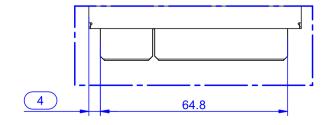
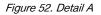


Figure 51. Front view









15.3 OPTION OF ADDING KEYING SCREW

A thread added to the side of the PET2000-12-074xA allows the user to add a screw to prevent the PET2000-12-074xA from being inserted into systems using other card edge connector types with the same power supply width and height. In such case, systems using PET2000-12-074xA must have a slot of ø6 mm x 14 mm implemented to allow PET2000-12-074xA to be inserted. The maximum size of the screw head is ø6 mm and height 2.12 mm.

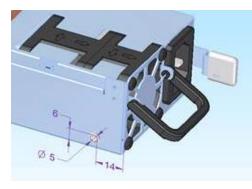


Figure 53. Polarizing screw

15.4 OUTPUT CONNECTOR PIN LOCATIONS

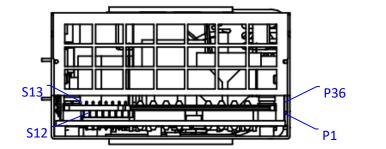


Figure 54. Rear view



Figure 55. Card edge PCB top view

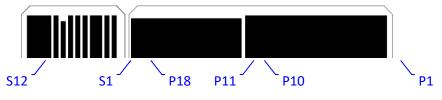


Figure 56. Card edge PCB bottom view



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16. CONNECTORS

PARAMETER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
AC inlet	PET2000-12-074xA : IEC 60320-C14 PET2000-12-074xAC : IEC 60320-C16 PET2000-12-074NAA: Anderson Saf-D-Grid®, P/N 2006G1				
AC cord requirement	Wire size	16			AWG
Output connector	36 Power- + 24 Signal-Pins PCB card edge				
	Manufacturer: FCI Electronics				
Mating output connector	Manufacturer P/N: 10130248-005LF (see <i>Figure 59</i> for option x)				
	Bel Power Solutions P/N: ZES.00678				

16.1 MATING OUTPUT CONNECTOR SPECIFICATION

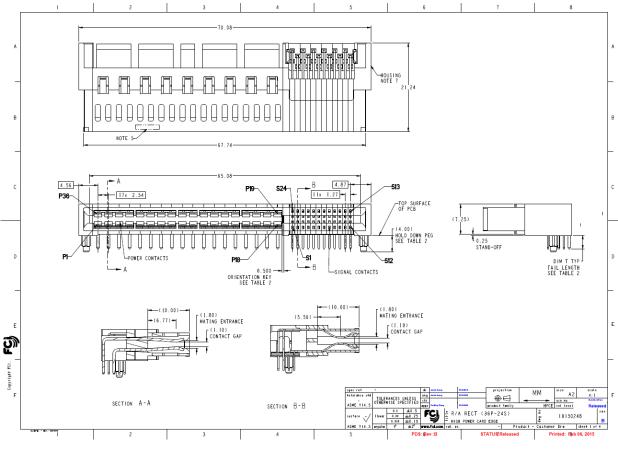


Figure 57. Mating connector drawing page 1



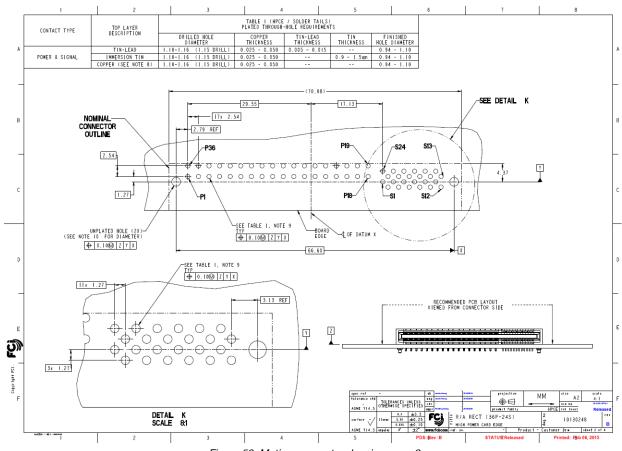


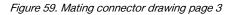
Figure 58. Mating connector drawing page 2



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HPCE PART NUMBER (TABLE 2) OR IENTATION KEY HOLD-DOWN OPTION DIM T TAIL LENGTH TAIL TYPE PART NUMBER ±0.25 SOLDER 10130248-001LF YES YES 2.60 10130248-002LF SOLDER TAIL YES 2.60 NO SOLDER TAIL NO 10130248-003LF NO 2.60 SOLDER TAIL 10130248-004LF YES NO 2.60 SOLDER TAIL 10130248-005LF YES YES 3.25 SOLDER TAIL YES 10130248-006LF NO 3.25 HOLD-DOWN OPTION SOLDER TAIL 10130248-007LF NO NO 3.25 10130248-008LF SOLDER TAIL YES NO 3.25 SOLDER TAIL YES YES 4.05 10130248-009LF SOLDER TAIL NOTES: 10130248-010LF NO YES 4.05 I. CONNECTOR MATERIALS: 10130248-01LF SOLDER TAIL NO 4.05 NO HIGH TEMPERATURE THERMAL PLASTIC, BLACK UL 94V-0 COMPLIANT HIGH PERFORMANCE COPPER ALLOY. HOUS ING : SOLDER TAIL 10130248-0121 F YES NO 4.05 CONTACTS: CONTACT FINISH REF. GS-12-604 SECTION 5.2. SOLDER TAIL 10130248-013LF YES YES 4.85 PRODUCT SPECIFICATION: GS-12-604. З. 10130248-014LF SOLDER TAIL NO YES 4.85 APPLICATION SPECIFICATION: GS-20-128. 10130248-015LF SOL DE R TAIL NO 4.85 PRODUCT MARKING (FCI - PART NUMBNER & DATE CODE) ON HOUSING IN AREA SHOWN NO (5.) PACKAGING MEETS FCI SPECIFICATION GS-14-937. 10130248-016LF SOLDER TAIL 6. YES 4.85 NO 7. HOUSING COMPONENT WILL WITHSTAND EXPOSURE TO 260°C PEAK TEMPERATURE FOR 60 SECONDS IN A CONVECTION, INFRA-RED, OR VAPOR PHASE REFLOW OVEN 8. 9. COPPER PLATING THICKNESS IN CENTER OF VIA-HOLE CAN BE NO MORE THAN 0.003 LESS THAN OTHER AREAS. ALL HOLE SIZES ARE FINISHED HOLE SIZES. MOUNTING HOLES ARE UNPLATED ∅ 2.10 +/- 0.1 FOR SOLDER TAILS ▲ ∅ 2.18 +/- 0.03 FOR LOW PLASTIC PEG INSERTION FORCE Ε apyright FCL. FC П. A ASSYMBOL WILL BE NEXT TO ANY DIMENSION , VIEW OR NOTE WHICH HAS BEEN MODIFIED WITH THE CURRENT DRAWING VERSION . spec ref tolerance std MM A2 4:1 TOLERANCES UNLESS OTHERWISE SPECIFIED ⊕⊖ ASME Y14.5 0.x ±0.5 0.x ±0.25 0.xx ±0.10 FCj R/A RECT (36P-24S) ou bap ······· -/ 10130248 HIGH POWER CARD EDGE sheet 4 SME YI omer Drw





с

16.2 MATING OUTPUT CONNECTOR SPECIFICATION

PIN	SIGNAL NAME	DESCRIPTION	MATING SEQUENCE ¹⁴
P1 ~ P10	GND		
P29 ~ P36	GND	Power and signal ground (return)	1
P11 ~ P18	V1		0
P19 ~ P28	V1	+12 VDC main output	2
S1	A0	120 address solation input	2
S2	A1	I ² C address selection input	2
S3, S4	VSB	+12 V Standby positive output (as pins S3, S4)	2
S 5	HOTSTANDBYEN_H	Hot standby enable signal, active-high	2
S 6	ISHARE	Analog current share bus	2
S7	Reserved	For future use, do not connect	2
S8	PRESENT_L	Power supply seated, active-low	3
S 9	A2	I ² C address selection input	2
S10 ~ S15	GND	Power and signal ground (return)	2
S16	PWOK_H	Power OK signal output, active-high	2
S17	V1_SENSE	Main output positive sense	2
S18	V1_SENSE_R	Main output negative sense	2
S19	SMB_ALERT_L	SMB Alert signal output, active-low	2
S20	PSON_L	Power supply on input, active-low	3
S21, S22	VSB	+12 V Standby positive output (as pins S3, S4)	2
S23	SCL	I ² C clock signal line	2
S24	SDA	I ² C data signal line	2

Table 5. Output connector pin assignment

¹⁴ 1 = First, 3 = Last, given by different card edge finger pin lengths and mating connector pin arrangement



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17. ACCESSORIES

ITEM	DESCRIPTION	ORDERING PART NUMBER	SOURCE
	I²C Utility Windows XP/Vista/7 compatible GUI to program, control and monitor Front-End power supplies (and other I ² C units)	ZS-00130	belfuse.com/power-solutions
	Evaluation Board Connector board to operate PET2000-12-074xA. Includes an on- board USB to I ² C converter (use I ² C Utility as desktop software).	YTM.00046	belfuse.com/power-solutions
Saf-D-Grid Plug	AC cable for PET2000-12-074NAA Anderson Saf-D-Grid® receptacle to IEC 60320-C20 plug, 14 AWG, 2 m, Anderson P/N 2052KH2	TBD	

18. REVISION HISTORY

DATE F	REVISION	DESCRIPTION OF CHANGE	ECO/MCO REFERENCE NO.
2019-Jun-19 A	4L	Page 4: Max. Output currents updated in the table	C94036

For more information on these products consult: tech.support@psbel.com

NUCLEAR AND MEDICAL APPLICATIONS - Products are not designed or intended for use as critical components in life support systems, equipment used in hazardous environments, or nuclear control systems.

TECHNICAL REVISIONS - The appearance of products, including safety agency certifications pictured on labels, may change depending on the date manufactured. Specifications are subject to change without notice.



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