



PFE3600-12-069RA

3600 W AC-DC Front-End Power Supply

PFE3600-12-069RA is a 3600 Watt AC/DC power-factor-corrected (PFC) and DC-DC power supply that converts standard AC mains power or high voltage DC bus voltages into a main output of 12 VDC for powering intermediate bus architectures (IBA) in high performance and reliability servers, routers, and network switches.

The PFE3600-12-069RA meets international safety standards and displays the CE-Mark for the European Low Voltage Directive (LVD).



Key Features & Benefits

- Best-in-class, Platinum efficiency
- Wide input voltage range: 90 – 300 VAC
- AC input with power factor correction
- DC input voltage range: 192 – 400 VDC
- Hot-plug capable
- Parallel operation with active current sharing thru analog bus
- Full digital controls for improved performance
- High density design: 37 W/in³
- Small form factor: 555 x 69 x 42 mm (21.85 x 2.72 x 1.65 in)
- I2C communication interface with Power Management Bus communication protocol for monitoring, control, and firmware update via bootloader
- Overtemperature, output overvoltage and overcurrent protection
- RoHS Compliant
- 2 Status LEDs: AC OK and DC OK with fault signaling
- US Patent Pending

Applications

- High Performance Servers
- Routers
- Switches



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1. ORDERING INFORMATION

PFE	3600	-	12	-	069	R	A	Option Code
Product Family	Power Level	Dash	V1 Output	Dash	Width	Airflow	Input	Blank: Standard model K: Screw for Key-in feature is installed.
PFE Front-Ends	3600 W		12 V		69 mm	R: Reversed ¹	A: AC	

¹ Front to Rear

2. OVERVIEW

The PFE3600-12-069RA is a fully DSP controlled, highly efficient front-end power supply. It incorporates resonant-soft-switching technology and interleaved power trains to reduce component stresses, providing increased system reliability and very high efficiency. With a wide input operating voltage range and minimal linear derating of output power with respect to ambient temperature, the PFE3600-12-069RA maximizes power availability in demanding server, switch, and router applications. The power supply is fan cooled and ideally suited for server integration with a matching airflow path.

The PFC stage is digitally controlled using a state-of-the-art digital signal processing algorithm to guarantee best efficiency and unity power factor over a wide operating range.

The DC-DC stage uses soft switching resonant techniques in conjunction with synchronous rectification. An active OR-ing device on the output ensures no reverse load current and renders the supply ideally suited for operation in redundant power systems. The always-on +12V standby output provides power to external power distribution and management controllers. Its protection with an active OR-ing device provides for maximum reliability.

Status information is provided with front-panel LEDs. In addition, the power supply can be monitored and controlled (i.e. fan speed setpoint) via I²C communication interface with Power Management Bus protocol. It allows full monitoring of the supply, including input and output voltage, current, power, and inside temperatures. The same I²C bus supports the bootloader to allow field update of the firmware in the DSP controllers.

Cooling is managed by a fan, controlled by the DSP controller. The fan speed is adjusted automatically depending on the actual power demand and supply temperature and can be overridden through the I²C bus.

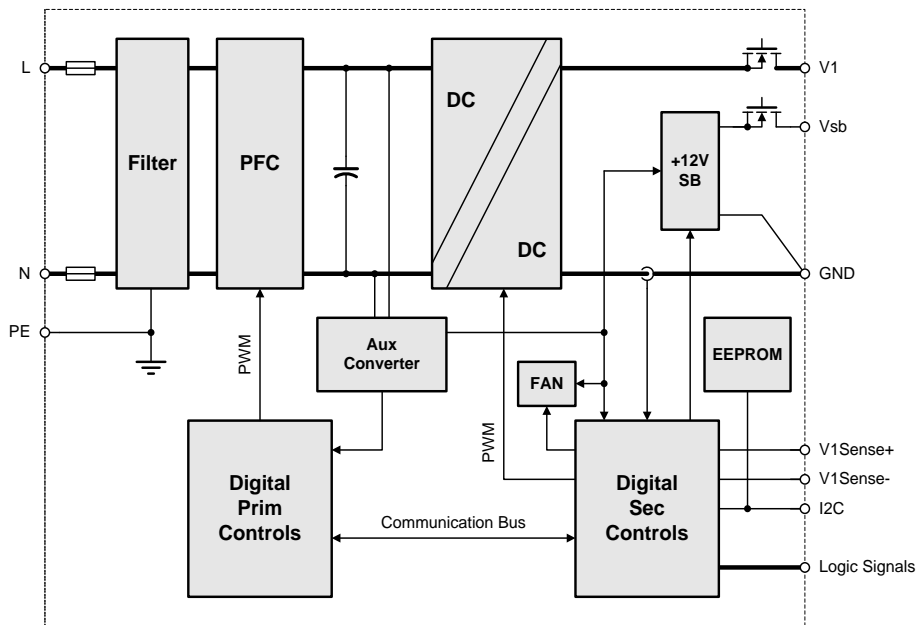


Figure 1. PFE3600-12-069RA Block Diagram

3. ABSOLUTE MAXIMUM RATINGS

Stresses in excess of the absolute maximum ratings may cause performance degradation, adversely affect long-term reliability, and cause permanent damage to the supply.

PARAMETER	CONDITIONS / DESCRIPTION	MIN	MAX	UNITS
$V_{i\ maxc}$	Maximum Input		300	VAC

4. INPUT

General Condition: $T_A = 0 \dots 45\ ^\circ\text{C}$ unless otherwise noted.

PARAMETER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT		
$V_{i\ nom}$	AC Nominal Input Voltage	100	230	277	VAC		
V_i	AC Input Voltage Ranges	Normal operating ($V_{i\ min}$ to $V_{i\ max}$)		300	VAC		
$V_{i\ nom\ DC}$	DC Nominal input voltage	240		380	VDC		
$V_{i\ DC}$	DC Input voltage ranges	Normal operating ($V_{i\ min}$ to $V_{i\ max}$)		400	VDC		
$V_{i\ red}$	Derated Input Voltage Range	See <i>Figure 20</i> and <i>Figure 33</i>		180	VAC		
$I_{i\ max}$	Max Input Current	$V_i > 200\ \text{VAC}, > 100\ \text{VAC}$		20	A_{rms}		
$I_{i\ p}$	Inrush Current Limitation	$V_{i\ min}$ to $V_{i\ max}, 0^\circ\ T_{NTC} = 25^\circ\text{C}$ (<i>Figure 5</i>)		50	A_p		
F_i	Input Frequency	47	50/60	63	Hz		
PF	Power Factor	$V_{i\ nom}, 50\text{Hz}, > 0.3\ I_{i\ nom}$		0.96	W/VA		
$V_{i\ on}$	Turn-on Input Voltage ²	Ramping up		85	88.5	90	VAC
$V_{i\ off}$	Turn-off Input Voltage ²	Ramping down		80	85	85	VAC
η	Efficiency without Fan	$V_{i\ nom}, 0.1 \cdot I_{i\ nom}, V_{x\ nom}, T_A = 25^\circ\text{C}$		90.0	91.85	%	
		$V_{i\ nom}, 0.2 \cdot I_{i\ nom}, V_{x\ nom}, T_A = 25^\circ\text{C}$		93.0	94.40		
		$V_{i\ nom}, 0.5 \cdot I_{i\ nom}, V_{x\ nom}, T_A = 25^\circ\text{C}$		94.5	94.95		
		$V_{i\ nom}, I_{i\ nom}, V_{x\ nom}, T_A = 25^\circ\text{C}$		92.0	93.0		
T_{hold}	Hold-up Time	After last AC zero point, $V_i > 10.8\ \text{V}$, V_{SB} within regulation, $V_i = 230\ \text{VAC}, P_{x\ nom}$		10		ms	

² The Front-End is provided with a minimum hysteresis of 3 V during turn-on and turn-off within the ranges

4.1 INPUT FUSE

Quick-acting 30 A input fuses (6.3 × 32 mm) in series with both the L- and N-line inside the power supply protect against severe defects. The fuses are not accessible from the outside and are therefore not serviceable parts.

4.2 INRUSH CURRENT

The AC-DC power supply exhibits an X capacitance of only 4.3 μF , resulting in a low and short peak current, when the supply is connected to the mains. The internal bulk capacitor will be charged through an NTC which will limit the inrush current.

NOTE:

Do not repeat plug-in / out operations below 90 sec interval time at maximum input, high temperature condition, or else the internal in-rush current limiting device (NTC) may not sufficiently cool down and excessive inrush current or component failure(s) may result.



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4.3 INPUT UNDER-VOLTAGE

If the RMS value of input voltage (either AC or DC) stays below the input undervoltage lockout threshold $V_{i on}$, the supply will be inhibited. Once the input voltage returns within the normal operating range, the supply will return to normal operation again.

4.4 POWER FACTOR CORRECTION

Power factor correction (PFC) (see *Figure 4*) is achieved by controlling the input current waveform synchronously with the input voltage. A fully digital controller is implemented giving outstanding PFC results over a wide input voltage and load ranges. The input current will follow the shape of the input voltage. If for instance the input voltage has a trapezoidal waveform, then the current will also show a trapezoidal waveform. At DC input voltage, the PFC is still in operation, but the input current will be DC in this case.

4.5 EFFICIENCY

The high efficiency (see *Figure 2*) is achieved by using state-of-the-art silicon power devices in conjunction with soft-transition topologies minimizing switching losses and a full digital control scheme. Synchronous rectifiers on the output reduce the losses in the high current output path. The rpm of the fan is digitally controlled to keep all components at an optimal operating temperature regardless of the ambient temperature and load conditions. *Figure 3* shows efficiency when input voltage is supplied from a high voltage DC source.

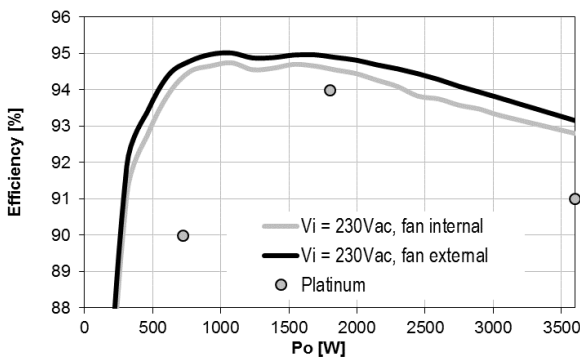


Figure 2. AC Input Efficiency vs. Load Current (ratio metric loading)

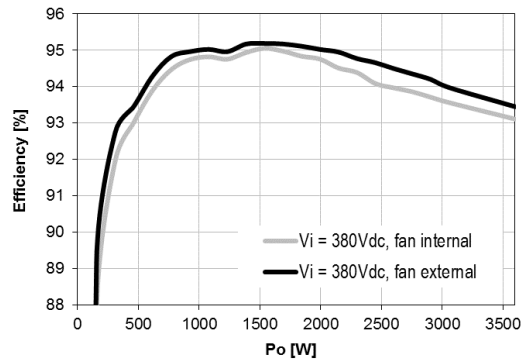


Figure 3. DC Input Efficiency vs. Load Current (ratio metric loading)

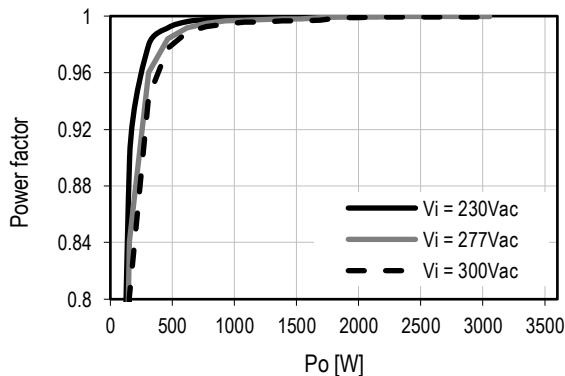


Figure 4. Power Factor vs. Load Current



Figure 5. Inrush Current, $V_{in} = 230 \text{ VAC}$, 0° phase angle
CH4: V_{in} (200 V/div), CH3: I_{in} (10 A/div)

5. OUTPUT

General Condition: $T_A = 0...45\text{ °C}$ unless otherwise noted.

PARAMETER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
Main Output V_1					
$V_1\text{ nom}$	Nominal Output Voltage	$0.5 \cdot I_1\text{ nom}$, $T_{\text{amb}} = 25\text{ °C}$	12.3		VDC
$V_1\text{ set}$	Output Setpoint Accuracy	-0.5		+0.5	% $V_1\text{ nom}$
$dV_1\text{ tot}$	Total Regulation	$V_1\text{ min}$ to $V_1\text{ max}$, 0 to 100% $I_1\text{ nom}$, $T_a\text{ min}$ to $T_a\text{ max}$		-1	+1 % $V_1\text{ nom}$
$P_1\text{ nom}$	Nominal Output Power	$V_1 = 12.3\text{ VDC}$, $V_{\text{in}} < 180\text{ VAC}$	1400		W
$I_1\text{ nom}$	Nominal Output Current	$V_1 = 12.3\text{ VDC}$, $V_{\text{in}} < 180\text{ VAC}$	114		ADC
$P_1\text{ nom}$	Nominal Output Power	$V_1 = 12.3\text{ VDC}$, $V_{\text{in}} > 180\text{ VAC}$	3600		W
$I_1\text{ nom}$	Nominal Output Current	$V_1 = 12.3\text{ VDC}$, $V_{\text{in}} > 180\text{ VAC}$	293		ADC
$I_1\text{ ol}$	Short time over load current	$V_1 = 12.3\text{ VDC}$, $V_{\text{in}} > 180\text{ VAC}$ $T_a\text{ min}$ to $T_a\text{ max}$, maximum duration 20 ms (See Section 5.2)		350	A
$V_1\text{ pp}$	Output Ripple Voltage	$V_1\text{ nom}$, $I_1\text{ nom}$, 20 MHz BW (See Section 5.1)		160	mVpp
$dV_1\text{ Load}$	Load Regulation	$V_1 = V_1\text{ nom}$, 0 - 100 % $I_1\text{ nom}$	170		mV
$dV_1\text{ Line}$	Line Regulation	$V_1 = V_1\text{ min}... V_1\text{ max}$	0		mV
$I_1\text{ ol lim}$	Current limitation	$V_1 < 180\text{ VAC}$, $T_a < 45\text{ °C}$	120	127	ADC
		$V_1 < 180\text{ VAC}$, $T_a = 55\text{ °C}$ ³	80	85	
		$V_1 > 180\text{ VAC}$, $T_a < 45\text{ °C}$	298	329	
		$V_1 > 180\text{ VAC}$, $T_a = 55\text{ °C}$ ³	200	220	
dI_{share}	Current Sharing	Deviation from $I_1\text{ tot} / N$, $I_1 > 30\%$ $I_1\text{ nom}$	-5%	+5%	A
dV_{dyn}	Dynamic Load Regulation	$\Delta I_1 = 50\%$ $I_1\text{ nom}$, $I_1 = 5 \dots 100\%$ $I_1\text{ nom}$, $dI_1/dt = 1\text{A}/\mu\text{s}$, $f_{\Delta I_1} = 0.05\dots 10\text{ kHz}$, Duty $\Delta I_1 = 10\dots 90\%$, recovery within 1% of V_1 final steady state	-0.6	+0.6	V
T_{rec}	Recovery Time			0.5	ms
$t_{\text{AC } V_1}$	Start-up Time from AC	$V_1 = 10.8\text{ VDC}$ (see Figure 7)		3	sec
$t_{V_1\text{ rise}}$	Rise Time	$V_1 = 10\dots 90\%$ $V_1\text{ nom}$ (see Figure 8)	2.5		ms
C_{Load}	Capacitive Loading	$T_a = 25\text{ °C}$		30000	μF

³ See

⁴ Figure 20 for linear derating > 45°C

Standby Output V_{SB}					
$V_{SB}\text{ nom}$	Nominal Output Voltage		12		VDC
$V_{SB}\text{ set}$	Output Setpoint Accuracy	$I_{SB}\text{ nom}$, $T_{\text{amb}} = 25\text{ °C}$	-0.5	+0.5	% $V_{SB}\text{ nom}$
$dV_{SB}\text{ tot}$	Total Regulation	$V_1\text{ min}$ to $V_1\text{ max}$, $I_{SB}\text{ nom}$, $T_a\text{ min}$ to $T_a\text{ max}$	-3	+3	% $V_{SB}\text{ nom}$
$P_{SB}\text{ nom}$	Nominal Output Power	$V_{SB} = 12\text{ VDC}$	60		W
$I_{SB}\text{ nom}$	Nominal Output Current	$V_{SB} = 12\text{ VDC}$	5		ADC
$V_{SB}\text{ pp}$	Output Ripple Voltage	$V_{SB}\text{ nom}$, $I_{SB}\text{ nom}$, 20 MHz BW (See Section 5.1)		300	mVpp
dV_{SB}	Droop	0 - 100 % $I_{SB}\text{ nom}$	400		mV
$I_{SB}\text{ lim}$	Current Limitation		6	9	ADC
$dV_{SB}\text{ dyn}$	Dynamic Load Regulation	$\Delta I_{SB} = 50\%$ $I_{SB}\text{ nom}$, $I_{SB} = 5 \dots 100\%$ $I_{SB}\text{ nom}$, $dI_{SB}/dt = 1\text{A}/\mu\text{s}$, $f_{\Delta I_{SB}} = 0.05\dots 10\text{ kHz}$, Duty $\Delta I_{SB} = 10\dots 90\%$, recovery within 1% of V_{SB} final steady state	-0.6	+0.6	$V_{SB}\text{ nom}$
T_{rec}	Recovery Time			0.5	ms
$t_{\text{AC } V_{SB}}$	Start-up Time from AC	$V_{SB} = 90\%$ $V_{SB}\text{ nom}$ (see Figure 7)		3	sec
$t_{V_{SB}\text{ rise}}$	Rise Time	$V_{SB} = 10\dots 90\%$ $V_{SB}\text{ nom}$ (see Figure 9)	10		ms
C_{Load}	Capacitive Loading	$T_{\text{amb}} = 25\text{ °C}$		3000	μF



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5.1 OUTPUT VOLTAGE RIPPLE

The internal output capacitance at the power supply output (behind OR-ing element) is minimized to prevent disturbances during hot plug. In order to provide low output ripple voltage in the application, external capacitors should be added close to the power supply output.

The setup of *Figure 6* has been used to evaluate suitable capacitor types. The capacitor combinations of Table 1 and Table 2 should be used to reduce the output ripple voltage.

The ripple voltage is measured with 20 MHz BWL, close to the external capacitors.

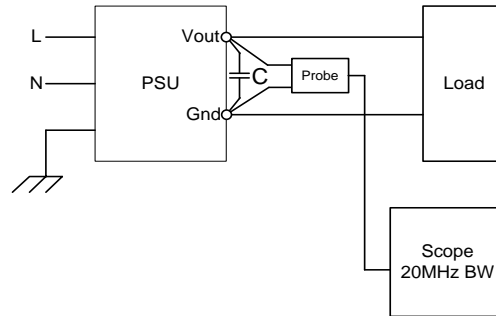


Figure 6 - Output Ripple Test Setup

NOTE: Care must be taken when using ceramic capacitors with a total capacitance of 1 μF to 50 μF on output V1, due to their high quality factor the output ripple voltage may be increased in certain frequency ranges due to resonance effects.

External Capacitor V1	dV1max	Unit
2Pcs 47 μF /16V/X5R/1210	160	mVpp
1Pcs 1000 μF /16V/Low ESR Aluminum/ ϕ 10x20	160	mVpp
1Pcs 270 μF /16V/Conductive Polymer/ ϕ 8x12	160	mVpp
2Pcs 47 μF /16V/X5R/1210 plus 1Pcs 270 μF Conductive Polymer OR 1Pcs 1000 μF Low ESR AlCap	90	mVpp

Table 1 - Suitable Capacitors for V1

External capacitor VSB	dVSBmax	Unit
1Pcs 10 μF /16 V/X7R/1206	300	mVpp

Table 2 - Suitable Capacitors for VSB

The output ripple voltage on VSB is influenced by the main output V1. Evaluating VSB output ripple must be done when maximum load is applied to V1.

5.2 SHORT TIME OVERLOAD

The main output has the capability to allow load current up to 20% above the nominal output current rating for a maximum duration of 20 ms. This allows the system to consume extended power for short time dynamic processes.

5.3 OUTPUT ISOLATION

Main and standby output and all signals are isolated from the chassis and protective earth connection, although the applied voltage must not exceed 100 Vpeak to prevent any damage of the supply.

Internal to the supply the main output ground, standby output ground and signal ground are interconnected through 10 Ω resistors to prevent any circulating current within the supply. In order to prevent any potential difference in outputs or signals within the application these 3 grounds must be directly interconnected at system level. See also section 14 for pins to be interconnected.

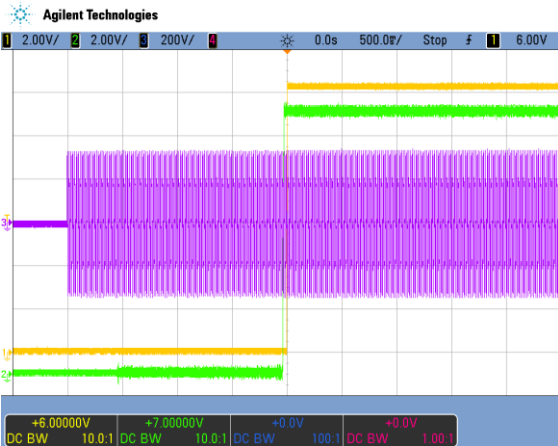


Figure 7. Turn-On AC Line 230 VAC, full load (500 ms/div)
CH1: V1 (2 V/div); CH2: VSB (2 V/div); CH3: Vin (200 V/div)

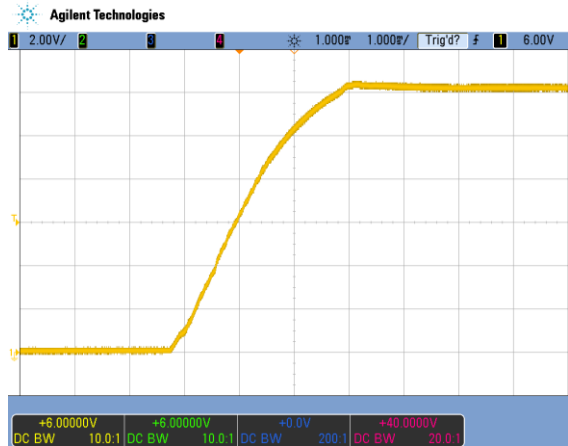


Figure 8. Turn-On AC Line 230 VAC, full load (1 ms/div)
CH1: V1 (2 V/div)

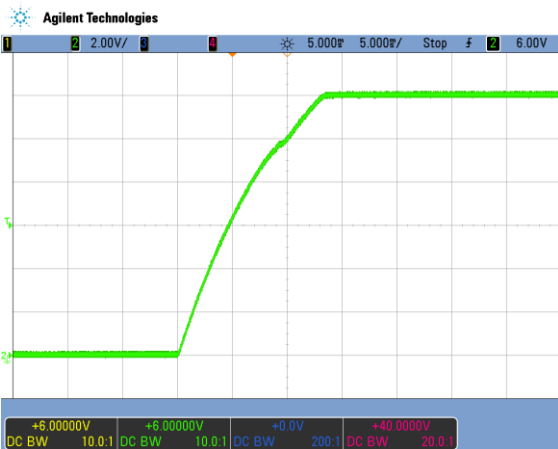


Figure 9. Turn-On AC Line 230 VAC, full load (5 ms/div)
CH2: VSB (2 V/div)



Figure 10. Turn-Off AC Line 230 VAC, full load (20 ms/div)
CH1: V1 (2 V/div); CH2: VSB (2 V/div); CH3: Vin (200 V/div)



Figure 11. Short circuit on V1 (50ms/div)
CH1: V1 (2V/div) CH2: VSB (2V/div) CH4: I1 (200A/div)

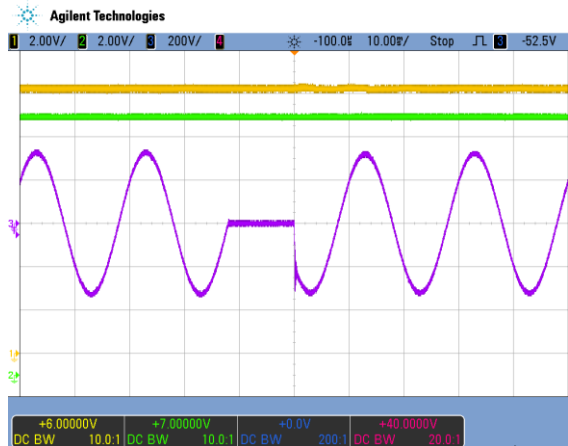


Figure 12. AC drop out 12ms (10ms/div)
CH1: V1 (2V/div) CH2: VSB (2V/div) CH3: Vin (200V/div)



Figure 13. AC drop out 40 ms, full load (20 ms/div)
CH1: V1 (2 V/div); CH2: VSB (2 V/div); CH3: Vin (200 V/div)

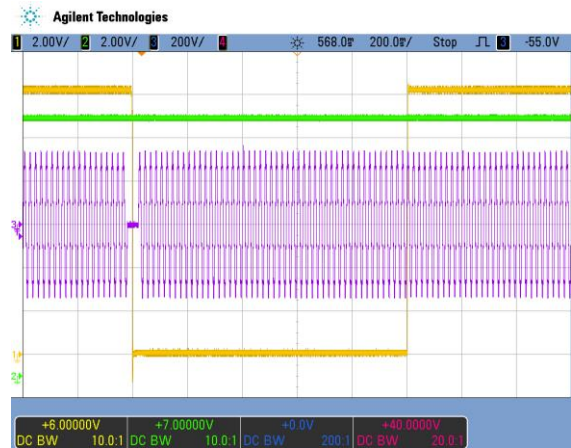


Figure 14. AC drop out 40 ms, full load (200 ms/div), V1 restart after 1 sec
CH1: V1 (5 V/div); CH2: VSB (2 V/div); CH3: I1 (200 V/div)

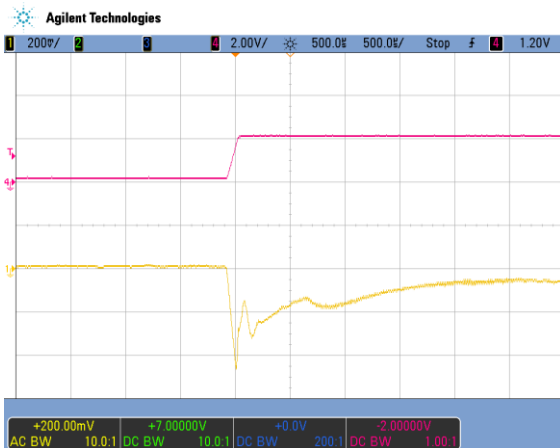


Figure 15. Load transient V1, 3 to 125 A (500 μs/div)
CH1: V1 (200 mV/div); CH4: I1 (100 A/div)

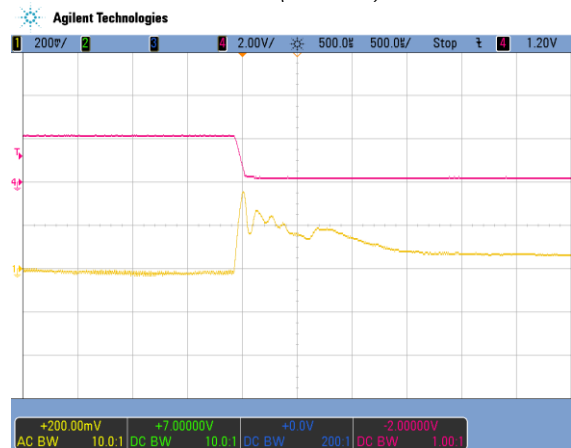


Figure 16. Load transient V1, 125 to 3 A (500 μs/div)
CH1: V1 (200 mV/div); CH4: I1 (100 A/div)

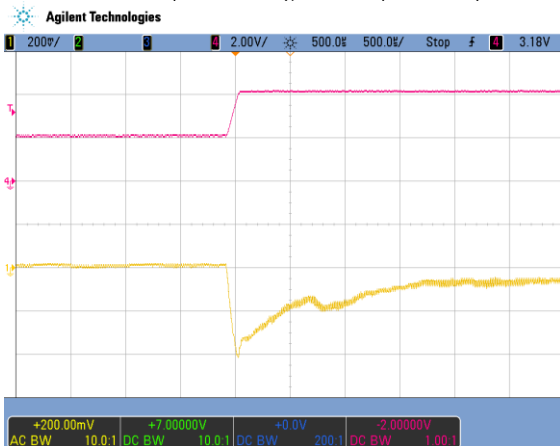


Figure 17. Load transient V1, 122 to 244 A (500 μs/div)
CH1: V1 (200 mV/div); CH4: I1 (100 A/div)

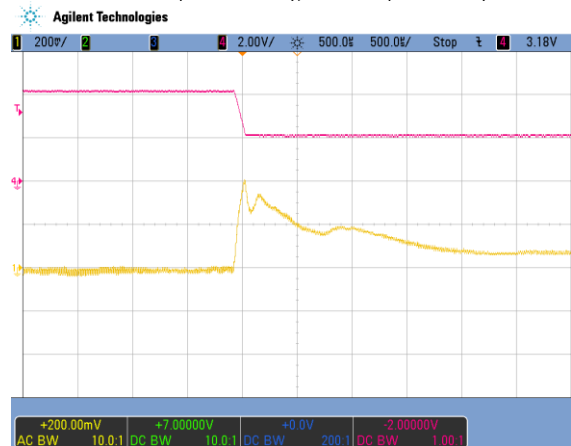


Figure 18. Load transient V1, 244 to 122 A (500 μs/div)
CH1: V1 (200 mV/div); CH4: I1 (100 A/div)

6. PROTECTION

PARAMETER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT		
F	Input Fuses (L+N)	Not user accessible, quick-acting (F)		30	A		
$V_{1\text{ ov}}$	OV Threshold V_1	13.6	14.2	14.8	VDC		
$t_{\text{bv } V_1}$	OV Latch Off Time V_1			1	ms		
$V_{\text{SB ov}}$	OV Threshold V_{SB}	13.3	13.9	14.5	VDC		
$t_{\text{bv } V_{\text{SB}}}$	OV Latch Off Time V_{SB}			1	ms		
$I_{1\text{ lim}}$	Current limitation	$V_1 < 180\text{ VAC}, T_a < 45^\circ\text{C}$	120	127	A		
		$V_1 < 180\text{ VAC}, T_a = 55^\circ\text{C}^4$	80	85			
		$V_1 > 180\text{ VAC}, T_a < 45^\circ\text{C}$	298	329			
		$V_1 > 180\text{ VAC}, T_a = 55^\circ\text{C}^4$	200	220			
$t_{V_1\text{ lim}}$	Current limit blanking time	Time to latch off when in over current		20	22	24	ms
$t_{V_1\text{ ol lim}}$	Current limit during short time overload V_1	Maximum duration 20 ms		328	336	344	A
$I_{1\text{ SC}}$	Max Short Circuit Current V_1	$V_1 < 3\text{ V}$			420 ⁵	A	
$t_{V_1\text{ SC off}}$	Short circuit latch off time	Time to latch off when in short circuit			10	ms	
$I_{\text{SB lim}}$	Current limitation V_{SB}	6		9	A		
$t_{\text{SB lim}}$	Current limit blanking time	Time to hit hiccup when in over current			1	ms	
T_{SD}	Over temperature on critical points	Inlet Ambient Temperature		60	°C		
		PFC Primary Heatsink Temperature		80			
		Secondary Sync Mosfet Temperature		120			
		Secondary OR-ing Mosfet Temperature		125			

⁵ See

⁶ Figure 20 for linear derating > 45°

⁷ Limit set don't include effects of main output capacitive discharge.

6.1 AUTOMATIC RETRY

For all fault conditions except current limitation on Standby output, the supply will shut down for 10 sec and restart automatically. The supply will auto-restart from a fault up to 5 times, after that it will latch off. The latch and restart counter can be cleared by recycling the input voltage or the PSON_L input. A failure on the Standby output will shut down both Main and Standby outputs. A failure on the Main output will shut down only the Main output, while Standby continues to operate.

6.2 OVERVOLTAGE PROTECTION

The PFE front-ends provide a fixed threshold overvoltage (OV) protection implemented with a HW comparator. Once an OV condition has been triggered, the supply will shut down and latch the fault condition.

6.3 UNDERVOLTAGE DETECTION

Both main and standby outputs are monitored. LED and PWOK_L pin signal if the output voltage exceeds $\pm 7\%$ of its nominal voltage.

Output undervoltage protection is provided on both outputs. When either V1 or VSB falls below 93% of its nominal voltage, the output is inhibited.

6.4 CURRENT LIMITATION

MAIN OUTPUT

Two different over current protection features are implemented on the main output.

A static over current protection will shut down the output, if the output current does exceed $I_{V1 \text{ lim}}$ for more than 20 ms.

If the output current is increased slowly this **protection will shut down the supply.**

The main output current limitation level $I_{V1 \text{ lim}}$ will decrease if the ambient (inlet) temperature increases beyond 45 °C (see

Figure 20). Note that the actual current limitation on V1 will kick in at a current level approximately 20 A higher than what is shown in

Figure 20 (see also Section 9 for additional information).

The 2nd protection is a substantially rectangular output characteristic controlled by a software feedback loop. This protects the power supply and system during the 20 ms blanking time of the static over current protection. If the output current is rising fast and reaches $I_{V1 \text{ ol lim}}$, the supply will immediately reduce its output voltage to prevent the output current from exceeding $I_{V1 \text{ ol lim}}$. When the output current is reduced below $I_{V1 \text{ ol lim}}$, the output voltage will return to its nominal value.

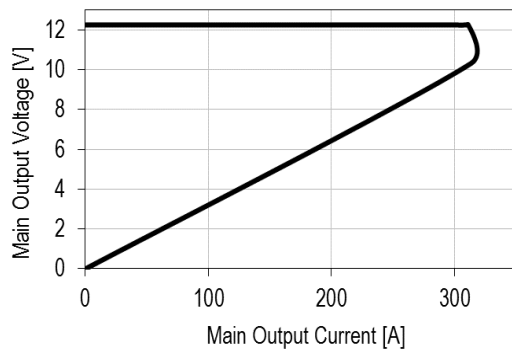


Figure 19. Current Limitation on V1 (Vi = 230VAC)

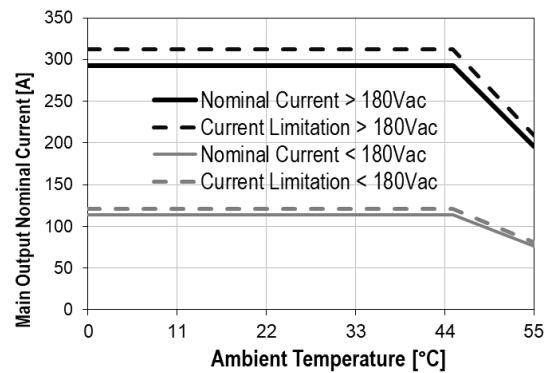


Figure 20. Derating on V1 vs. Ta

STANDBY OUTPUT

On the standby output a hiccup type over current protection is implemented. This protection will shut down the standby output immediately when standby current reaches or exceeds $I_{VSB \text{ lim}}$. After an off-time of 1s the output automatically tries to restart. If the overload condition is removed the output voltage will reach again its nominal value. At continuous overload condition the output will repeatedly trying to restart with 1s intervals.

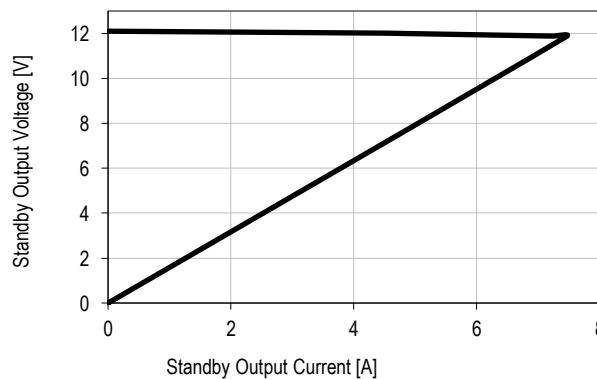


Figure 21. Current Limitation on Vsb

7. MONITORING

PARAMETER		DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
$V_{i\ mon}$	Input RMS Voltage	$V_{i\ min} \leq V_i \leq V_{i\ max}$	-2.5		+2.5	%
$I_{i\ mon}$	Input RMS Current	$I_i > 4\ A_{rms}$	-5		+5	%
		$I_i \leq 4\ A_{rms}$	-0.2		+0.2	A_{rms}
$P_{i\ mon}$	True Input Power	$P_i > 800\ W$	-5		+5	%
		$P_i \leq 800\ W$	-35		+35	W
$E_{i\ mon}$	Total Input Energy	$P_i > 800\ W$	-5		+5	%
		$P_i \leq 800\ W$	-35		+35	Wh
$V_{i\ mon}$	V ₁ Voltage		-2		+2	%
$I_{i\ mon}$	V ₁ Current	$I_{i1} > 30\ A$	-2		+2	%
		$I_{i1} \leq 30\ A$	-0.6		+0.6	A
$P_{o\ nom}$	Total Output Power	$P_o > 250\ W$	-5		+5	%
		$P_o \leq 250\ W$	-10		+10	W
$E_{o\ mon}$	Total Output Energy	$P_o > 250\ W$	-5		+5	%
		$P_o \leq 250\ W$	-10		+10	Wh
$V_{SB\ mon}$	Standby Voltage		-2		+2	%
$I_{SB\ mon}$	Standby Current	$I_{SB} \leq I_{SB\ nom}$	-0.3		+0.3	A

8. SIGNALING AND CONTROL

8.1 ELECTRICAL CHARACTERISTICS

PARAMETER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
PSKILL / PSON_L inputs					
V_L	Input low level voltage	-0.2		0.8	V
V_H	Input high level voltage	2.0		3.6	V
$I_{L, H}$	Maximum input sink or source current	0		1	mA
$R_{puPSKILL}$	Internal pull up resistor on PSKILL		10		k Ω
R_{puPSON_L}	Internal pull up resistor on PSON_L		10		k Ω
PWOK_L output					
V_{OL}	Output low level voltage	$I_{sink} < 4 \text{ mA}$	-0.2	0.4	V
V_{puPWOK_L}	External pull up voltage			12	V
R_{puPWOK_L}	Recommended external pull up resistor on PWOK_L at $V_{puPWOK_L} = 3.3 \text{ V}$		10		k Ω
<i>Low level output</i>	All outputs are turned on and within regulation				
<i>High level output</i>	In standby mode or V_1/V_{SB} have triggered a fault condition				
INOK_L output					
V_{OL}	Output low level voltage	$I_{sink} < 4 \text{ mA}$	-0.2	0.4	V
V_{puINOK_L}	External pull up voltage			12	V
R_{puINOK_L}	Recommended external pull up resistor on INOK_L at $V_{puINOK_L} = 3.3 \text{ V}$		10		k Ω
<i>Low level output</i>	Input voltage is within range for PSU to operate				
<i>High level output</i>	Input voltage is not within range for PSU to operate				
SMB_ALERT_L output					
V_{OL}	Output low level voltage	$I_{sink} < 4 \text{ mA}$	-0.2	0.4	V
$V_{puSMB_ALERT_L}$	External pull up voltage			12	V
$R_{puSMB_ALERT_L}$	Recommended external pull up resistor on SMB_ALERT_L at $V_{puSMB_ALERT_L} = 3.3 \text{ V}$		10		k Ω
<i>Low level output</i>	PSU in warning or failure condition				
<i>High level output</i>	PSU is ok				

8.2 INTERFACING WITH SIGNALS

A 15 V zener diode is added on all signal pins versus signal ground SGND to protect internal circuits from negative and high positive voltage. Signal pins of several supplies running in parallel can be interconnected directly. A supply having no input power will not affect the signals of the paralleled supplies.

ISHARE pins must be interconnected without any additional components. This in-/output also has a 15 V zener diode as a protection device and is disconnected from internal circuits when the power supply is switched off.

8.3 FRONT LEDs

The front-end has 2 front LEDs showing the status of the supply. LED number one is green and indicates AC power is on or off, while LED number two is bi-colored: green and yellow and indicates DC power presence or fault situations. For the position of the LEDs see **Table 3** listing the different LED status.

OPERATING CONDITION	LED SIGNALING
AC LED	
AC Line within range	Solid Green
AC Line UV condition	Off
DC LED*	
Normal Operation	Solid Green
PSON_L High	Blinking Yellow (1:1)
V_I or V_{SB} out of regulation	Solid Yellow
Over temperature shutdown	
Output over voltage shutdown (V_I or V_{SB})	
Output under voltage shutdown (V_I or V_{SB})	
Output over current shutdown (V_I or V_{SB})	Blinking Yellow/Green (2:1)
Over temperature warning	
Minor fan regulation error (>5%, <15%)	Blinking Yellow/Green (1:1)

* The order of the criteria in the table corresponds to the testing precedence in the controller.

Table 3. LED Status

8.4 PRESENT_L

The PRESENT_L is normally a trailing pin within the connector and will contact only once all other connector contacts are closed. This active-low pin is used to indicate to a power distribution unit controller that a supply is plugged in. The maximum sink current on PRESENT_L pin should not exceed 10 mA.

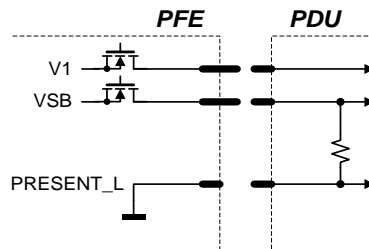


Figure 22. PRESENT_L signal pin

8.5 PSKILL INPUT

The PSKILL input is an active-low and normally a trailing pin in the connector and is used to disconnect the main output as soon as the power supply is being plugged out. This pin should be connected to SGND in the power distribution unit. The standby output will remain on regardless of the PSKILL input state.

8.6 AC TURN-ON / DROP-OUTS / INOK_L

The power supply will automatically turn-on when connected to the AC line under the condition that the PSON_L signal is pulled low and the AC line is within range. The INOK_L is an open collector output that requires an external pull-up to a maximum of 12V indicating whether the input is within the range the power supply can use and turn on. The INOK_L signal is active-low. The timing diagram is shown in **Figure 23** and referenced in **Table 4**.

OPERATING CONDITION		MIN	MAX	UNIT
$t_{AC VSB}$	AC Line to 90% V_{SB}		3	sec
$t_{AC V1}$	AC Line to 90% V_1		3	sec
$t_{INOK_L\ on1}$	INOK_L signal on delay (start-up)		1800	ms
$t_{INOK_L\ on2}$	INOK_L signal on delay (dips)	0	100	ms
$t_{V1\ holdup}$	Effective V_1 holdup time	12	300	ms
$t_{VSB\ holdup}$	Effective V_{SB} holdup time	40	300	ms
$t_{INOK_L\ V1}$	INOK_L to V_1 holdup	7		ms
$t_{INOK_L\ VSB}$	INOK_L to V_{SB} holdup	27		ms
$t_{V1\ off}$	Minimum V_1 off time	1000	1200	ms
$t_{VSB\ off}$	Minimum V_{SB} off time	1000	1200	ms
$t_{V1\ dropout}$	Minimum V_1 dropout time	12		ms
$t_{VSB\ dropout}$	Minimum V_{SB} dropout time	40		ms

Table 4. AC Turn-on / Dip Timing

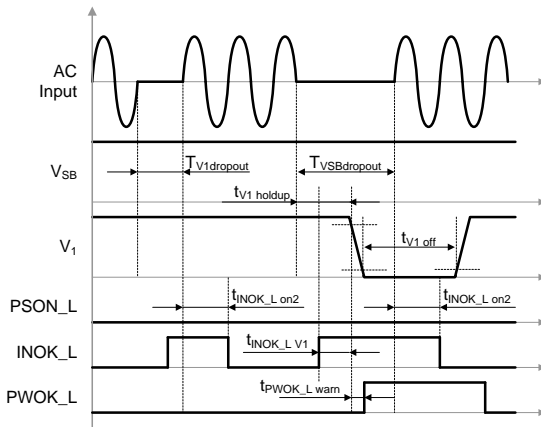


Figure 24. AC short dips

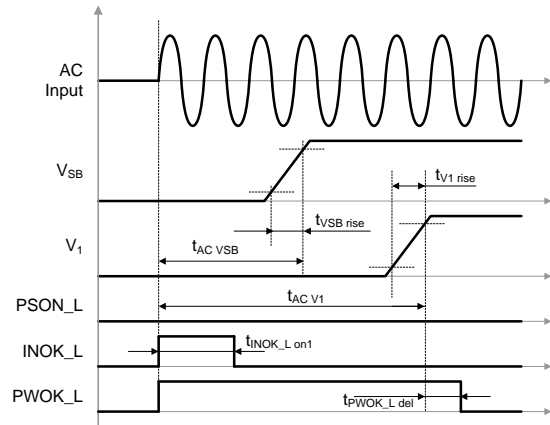


Figure 23. AC turn-on timing

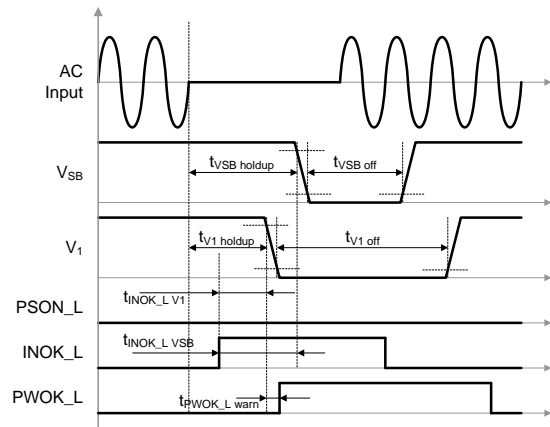


Figure 25. AC long dips

8.7 PSON_L INPUT

The PSON_L is an internally pulled-up (3.3 V) input signal to enable / disable the main output V1 of the front-end. This active-low pin is also used to clear any latched fault condition. The timing diagram is given in Figure 26 and the parameters in Table 5.

OPERATING CONDITION		MIN	MAX	UNIT
$t_{PSON_L\ V1\ on}$	PSON_L to V_1 delay (on)	190	220	ms
$t_{PSON_L\ V1\ off}$	PSON_L to V_1 delay (off)	0	100	ms

Table 5. PSON_L timing

8.8 PWOK_L SIGNAL

The PWOK_L is an open collector output that requires an external pull-up to a maximum of 12 V indicating whether both VSB and V1 outputs are within regulation. This pin is active-low. The timing diagram is shown in Figure 26 and referenced in Table 6.

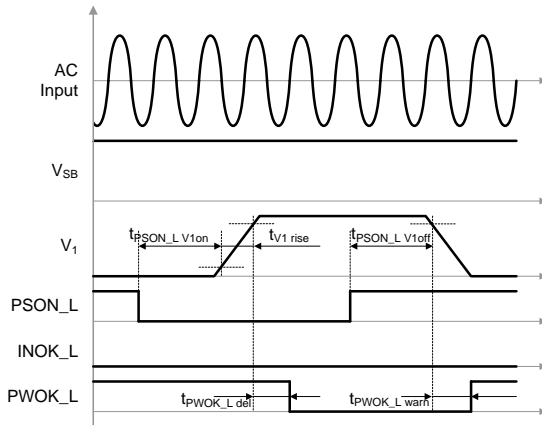


Figure 26. PSON_L turn-on/off timing

OPERATING CONDITION		MIN	MAX	UNIT
tPWOK_L del	V _I to PWOK_L delay (on)	250	350	ms
tPWOK_L warn	V _I to PWOK_L delay (off)	0	5	ms

Table 6. PWOK_L timing

8.9 CURRENT SHARE

The PFE front-ends have an active current share scheme implemented for V1. All the ISHARE current share pins need to be interconnected in order to activate the sharing function. If a supply has an internal fault or is not turned on, it will disconnect its ISHARE pin from the share bus. This will prevent dragging the output down (or up) in such cases.

The current share function uses an analog bus. The controller implements a Master/Slave current share function. The power supply providing the largest current among the group is automatically the Master. The other supplies will operate as Slaves and increase their output current to a value close to the Master by slightly increasing their output voltage. The voltage increase is limited to +250 mV.

The standby output uses a passive current share method (droop output voltage characteristic).

No of paralleled PSUs	Maximum available power on main 12 V without redundancy	Maximum available power on main 12 V with n+1 redundancy	Maximum available power on standby output
1	3600 W	-	60 W
2	7020 W	3600 W	60 W
3	10440 W	7020 W	60 W
4	13860 W	10440 W	60 W
5	17280 W	13860 W	60 W
6	20700 W	17280 W	60 W

Table 7. Power available when PSU in redundant operation

8.10 SENSE INPUTS

Main output has sense lines implemented to compensate for voltage drop on load wires. The maximum allowed voltage drop is 200 mV on the positive rail and 100 mV on the PGND rail.

With open sense inputs the main output voltage will rise by 250 mV. Therefore, if not used, these inputs should be connected to the power output and PGND close to the power supply connector. The sense inputs are protected against short circuit. In this case the power supply will shut down.



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8.11 I2C / POWER MANAGEMENT BUS COMMUNICATION

The interface driver in the PFE supply is referenced to the SGND. The PFE supply is a communication slave device only; it never initiates messages on the I2C bus by itself. The communication bus voltage and timing is defined in **Table 8** and further characterized through:

- There are 100 kΩ internal pull-up resistors
- The SDA/SCL IOs must be pull-up externally to 3.3 ± 0.3 V
- Pull-up resistor should be 2 – 5 kΩ to ensure SMBUS compliant signal rise times
- I2C clock speed up to 100 kbps
- Clock stretching limited to 1 ms
- SCL low time-out of >25 ms with recovery within 10 ms
- Recognizes any time Start/Stop bus conditions

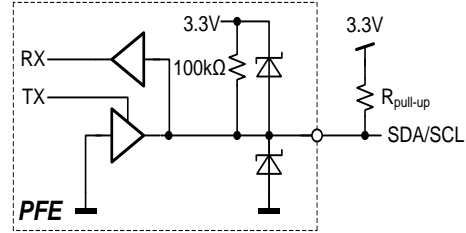


Figure 27. Physical layer of communication interface

The SMB_ALERT_L signal indicates that the power supply is experiencing a problem that the system agent should investigate. This is a logical OR of the Shutdown and Warning events. Communication to the DSP or the EEPROM will be possible as long as the input AC (DC) voltage is provided. If no AC (DC) is present, communication to the unit is possible as long as it is connected to a live VSB output (provided e.g. by the redundant unit). If only V1 is provided, communication is not possible.

PARAMETER	DESCRIPTION	CONDITION	MIN	MAX	UNIT
V_L	Input low voltage		-0.2	0.4	V
V_H	Input high voltage		2.1	3.6	V
V_{hys}	Input hysteresis		0.15		V
V_{OL}	Output low voltage	4 mA sink current	0	0.4	V
t_r	Rise time for SDA and SCL		$20+0.1C_b^*$	300	ns
t_{of}	Output fall time $V_{iHmin} \rightarrow V_{iLmax}$	$10\text{ pF} < C_b^* < 400\text{ pF}$	$20+0.1C_b^*$	250	ns
I_i	Input current SCL/SDA	$0.1\text{ VDD} < V_i < 0.9\text{ VDD}$	-10	10	μA
C_i	Capacitance for each SCL/SDA			10	pF
f_{SCL}	SCL clock frequency		0	100	kHz
R_{pu}	External pull-up resistor	$f_{SCL} \leq 100\text{ kHz}$		$1000\text{ ns} / C_b^*$	Ω
t_{HDSTA}	Hold time (repeated) START	$f_{SCL} \leq 100\text{ kHz}$	4.0		μs
t_{LOW}	Low period of the SCL clock	$f_{SCL} \leq 100\text{ kHz}$	4.7		μs
t_{HIGH}	High period of the SCL clock	$f_{SCL} \leq 100\text{ kHz}$	4.0		μs
t_{SUSTA}	Setup time for a repeated START	$f_{SCL} \leq 100\text{ kHz}$	4.7		μs
t_{HDDAT}	Data hold time	$f_{SCL} \leq 100\text{ kHz}$	0	3.45	μs
t_{SUDAT}	Data setup time	$f_{SCL} \leq 100\text{ kHz}$	250		ns
t_{SUSTO}	Setup time for STOP condition	$f_{SCL} \leq 100\text{ kHz}$	4.0		μs
t_{BUF}	Bus free time between STOP and START	$f_{SCL} \leq 100\text{ kHz}$	4.7		μs
EEPROM_WP					
V_L	Input low voltage		-0.2	0.4	V
V_H	Input high voltage		2.1	3.6	V
I_i	Input sink or source current		-1	1	mA
R_{pu}	Internal pull-up resistor to 3.3V			10k	Ω

* C_b = Capacitance of bus line in pF, typically in the range of 10...400 pF

Table 8. I2C / SMBus Specification

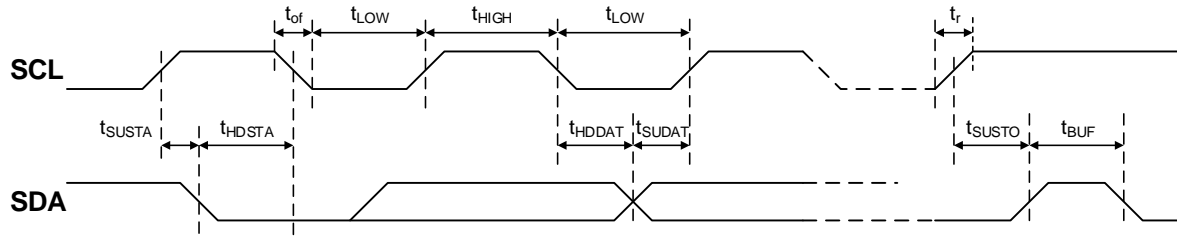


Figure 28. I²C / SMBus Timing

8.12 ADDRESS

The supply supports Power Management Bus communication protocol, address for Power Management Bus communication is at fixed to 0x20. The EEPROM is at fixed address = 0xA0.

8.13 CONTROLLER AND EEPROM ACCESS

The controller and the EEPROM in the power supply share the same I²C bus physical layer (see Figure 29). In order to write to the EEPROM, the write protection needs to be disabled by setting EEPROM_WP input correctly. If EEPROM_WP is High, write is not allowed to the EEPROM and if Low, write is allowed. The EEPROM provides 2k bytes of user memory. None of the bytes are used for the operation of the power supply.

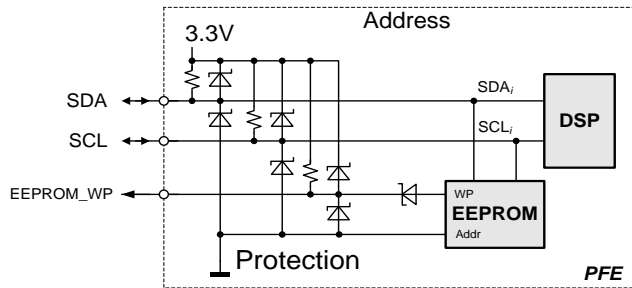


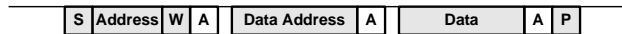
Figure 29. I²C Bus to DSP and EEPROM

8.14 EEPROM PROTOCOL

The EEPROM follows the industry communication protocols used for this type of device. Even though page write / read commands are defined, it is recommended to use the single byte write / read commands.

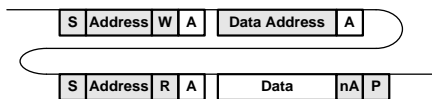
WRITE

The write command follows the SMBus 1.1 Write Byte protocol. After the device address with the write bit cleared a first byte with the data address to write to is sent followed by the data byte and the STOP condition. A new START condition on the bus should only occur after 5ms of the last STOP condition to allow the EEPROM to write the data into its memory.



READ

The read command follows the SMBus 1.1 Read Byte protocol. After the device address with the write bit cleared the data address byte is sent followed by a repeated start, the device address and the read bit set. The EEPROM will respond with the data byte at the specified location.



8.15 POWER MANAGEMENT BUS PROTOCOL

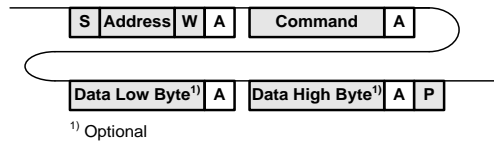
The Power Management Bus is an open standard protocol that defines means of communicating with power conversion and other devices. For more information, please see the System Management Interface Forum web site at: www.powerSIG.org.

Power Management Bus command codes are not register addresses. They describe a specific command to be executed. PFE3600-12-069RA supply supports the following basic command structures:

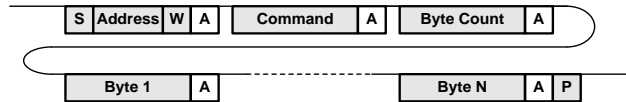
- Clock stretching limited to 1 ms
- SCL low time-out of >25 ms with recovery within 10 ms
- Recognized any time Start/Stop bus conditions

WRITE

The write protocol is the SMBus 1.1 Write Byte/Word protocol. Note that the write protocol may end after the command byte or after the first data byte (Byte command) or then after sending 2 data bytes (Word command).

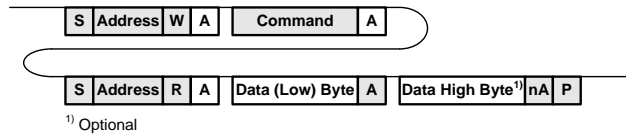


In addition, Block write commands are supported with a total maximum length of 255 bytes. See PFE3000-12-069RA Power Management Bus Communication Manual BCA.00070 for further information.

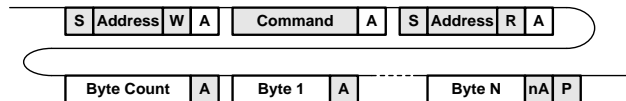


READ

The read protocol is the SMBus 1.1 Read Byte/Word protocol. Note that the read protocol may request a single byte or word.



In addition, Block read commands are supported with a total maximum length of 255 bytes. See PFE3000-12-069RA Power Management Bus Communication Manual BCA.00070 for further information.



8.16 GRAPHICAL USER INTERFACE

Bel Power Solutions I²C Utility provides a Windows® Vista/Win7/8 compatible graphical user interface allowing the programming and monitoring of the PFE3600-12-069RA Front-End. The utility can be downloaded on belfuse.com/power-solutions and supports the Power Management Bus protocol. The GUI allows automatic discovery of the units connected to the communication bus and will show them in the navigation tree. In the monitoring view the power supply can be controlled and monitored. If the GUI is used in conjunction with the PFE3600-12-069RA Evaluation Kit it is also possible to control the PSON_L pin(s) of the power supply.

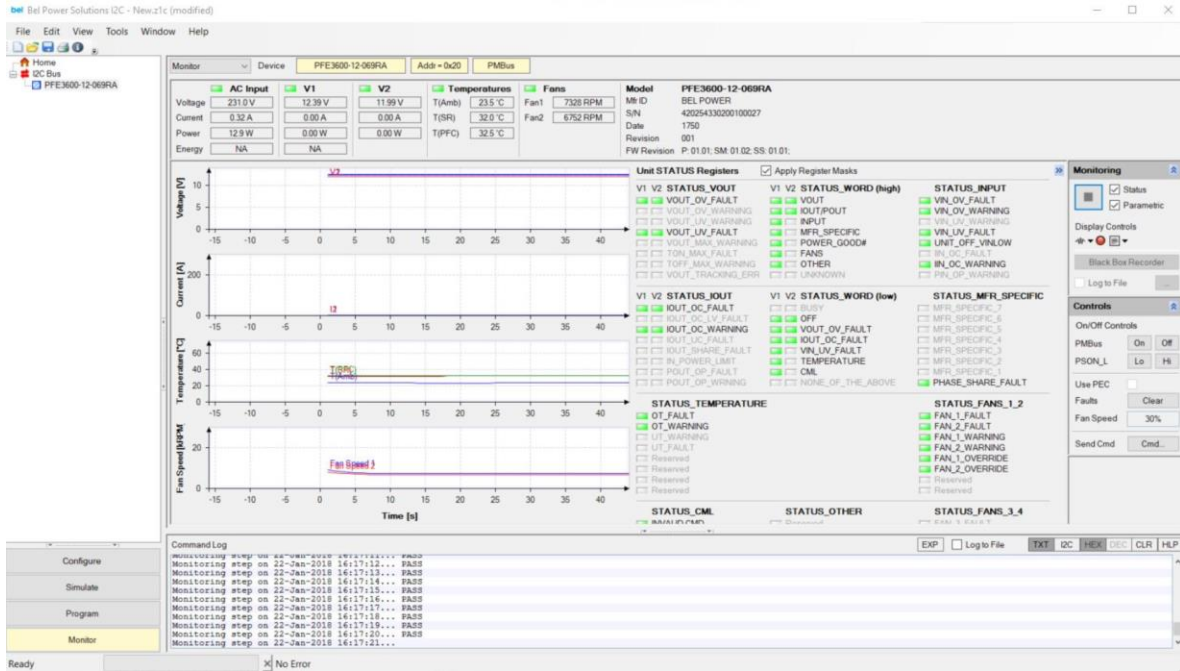


Figure 30. Monitoring dialog of the I2C Utility

9. TEMPERATURE AND FAN CONTROL

To achieve best cooling results sufficient airflow through the supply must be ensured. Do not block or obstruct the airflow at the rear of the supply by placing large objects directly at the output connector. The PFE3600-12-069RA is provided with a reverse airflow, which means the air enters through the front of the supply and leaves at the rear. PFE supplies have been designed for horizontal operation.

The fan inside of the supply is controlled by a microprocessor. The rpm of the fan is adjusted to ensure optimal supply cooling and is a function of output power and the inlet temperature.



Figure 31. Airflow Direction

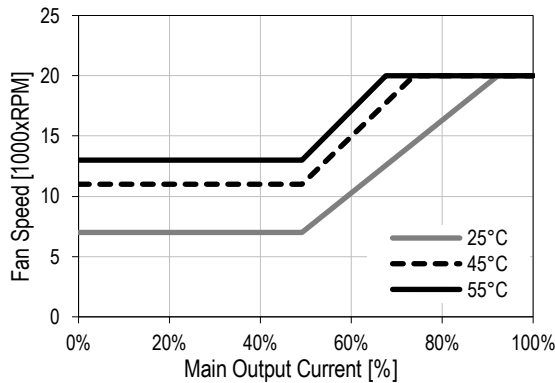


Figure 32. Fan speed vs. main output load for PFE3600-12-069RA

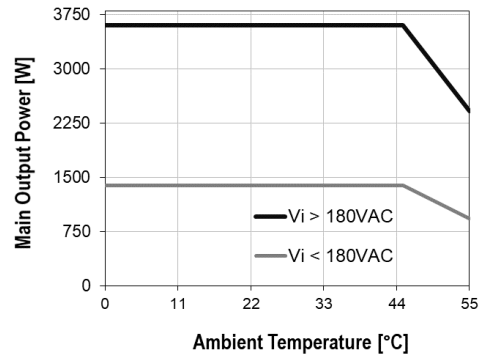


Figure 33. Thermal derating for PFE3600-12-069RA

10. ELECTROMAGNETIC COMPATIBILITY

10.1 IMMUNITY

NOTE: Most of the immunity requirements are derived from EN 55024:1998/A2:2003.

PARAMETER	DESCRIPTION / CONDITION	CRITERION
ESD Contact Discharge	IEC / EN 61000-4-2, ±8 kV, 25+25 discharges per test point (metallic case, LEDs, connector body)	A
ESD Air Discharge	IEC / EN 61000-4-2, ±15 kV, 25+25 discharges per test point (non-metallic user accessible surfaces)	A
Radiated Electromagnetic Field	IEC / EN 61000-4-3, 10 V/m, 1 kHz/80% Amplitude Modulation, 1 µs Pulse Modulation, 10 kHz...2 GHz	A
Burst	IEC / EN 61000-4-4, level 3 AC port ±2 kV, 1 minute DC port ±1 kV, 1 minute	A
Surge	IEC / EN 61000-4-5 Line to earth: level 3, ±2 kV Line to line: level 2, ±1 kV	A
RF Conducted Immunity	IEC/EN 61000-4-6, Level 3, 10 Vrms, CW, 0.1 ... 80 MHz	A
Voltage Dips and Interruptions	IEC/EN 61000-4-11 1: Vi 230Volts, 100% Load, Dip 100%, Duration 12ms 2: Vi 230Volts, 100% Load, Dip 100%, Duration < 150 ms 3: Vi 230Volts, 100% Load, Dip 100%, Duration > 150 ms	A V1: B, VSB: A B

10.2 EMISSION

PARAMETER	DESCRIPTION / CONDITION	CRITERION
Conducted Emission	EN55022 / CISPR 22: 0.15 ... 30 MHz, QP and AVG	Class A
Radiated Emission	EN55022 / CISPR 22: 30 MHz ... 1 GHz, QP	Class A
Harmonic Emissions	IEC61000-3-2, Vin = 115/230 VAC, 50 Hz, 100% Load	Class A
Acoustical Noise	Sound power statistical declaration (ISO 9296, ISO 7779, IS9295) @ 50% load	60 dBA
AC Flicker	IEC / EN 61000-3-3, d _{max} < 3.3%	PASS

11. SAFETY / APPROVALS

Maximum electric strength testing is performed in the factory according to IEC/EN 60950, and UL 60950. Input-to-output electric strength tests should not be repeated in the field. Bel Power Solutions will not honor any warranty claims resulting from electric strength field tests.

PARAMETER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
Agency Approvals	Approved to the latest edition of the following standards: <ul style="list-style-type: none"> • IEC60950-1 2nd edition (CB) • EN60950-1 2nd Edition (Nemko) • UL/CSA0950-1 2nd Edition (cCSAus) 				
Isolation Strength	Input (L/N) to case (PE) Input (L/N) to output Output to case (PE)			Basic Reinforced Functional	
d_c Creepage / Clearance	Primary (L/N) to protective earth (PE) Primary to secondary				
Electrical Strength Test	Input to case Input to output (tested by manufacturer only)	2121 4242			VDC

12. ENVIRONMENTAL

PARAMETER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
T_A Ambient Temperature	$V_{I \min}$ to $V_{I \max}$, $I_{I \text{ nom}}$, $I_{SB \text{ nom}}$ at 4000 m $V_{I \min}$ to $V_{I \max}$, $I_{I \text{ nom}}$, $I_{SB \text{ nom}}$ at 1800 m	0 0		+35 +45	°C
$T_{A \text{ ext}}$ Extended Temp. Range	Derated output (see Figure 20 and Figure 33) at 1800 m	+45		+55	°C
T_S Storage Temperature	Non-operational	-40		+70	°C
Altitude	Operational, above Sea Level (see derating)	-		4000	m
M_A Audible Noise	$V_{I \text{ nom}}$, 50% $I_{O \text{ nom}}$, $T_A = 25^\circ\text{C}$		60		dB(A)
Cooling	System Back Pressure			0.5	in-H ₂ O

13. MECHANICAL

PARAMETER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
Dimensions	Width		69		mm
	Height		42		mm
	Depth		555		mm
m Weight			2.60		kg

NOTE: A 3D step file of the power supply casing is available on request.

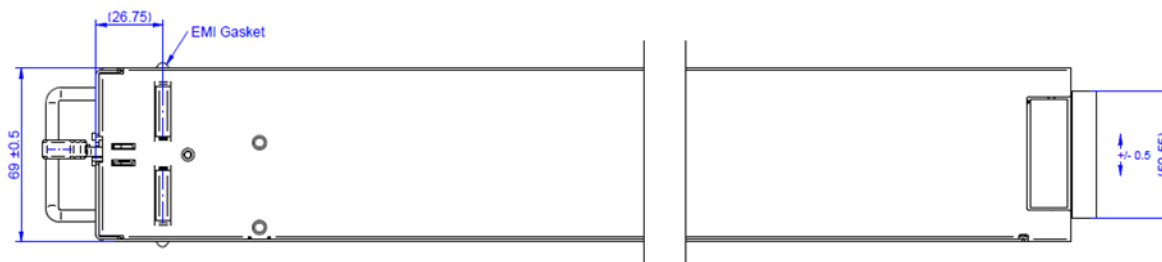


Figure 34. Bottom view



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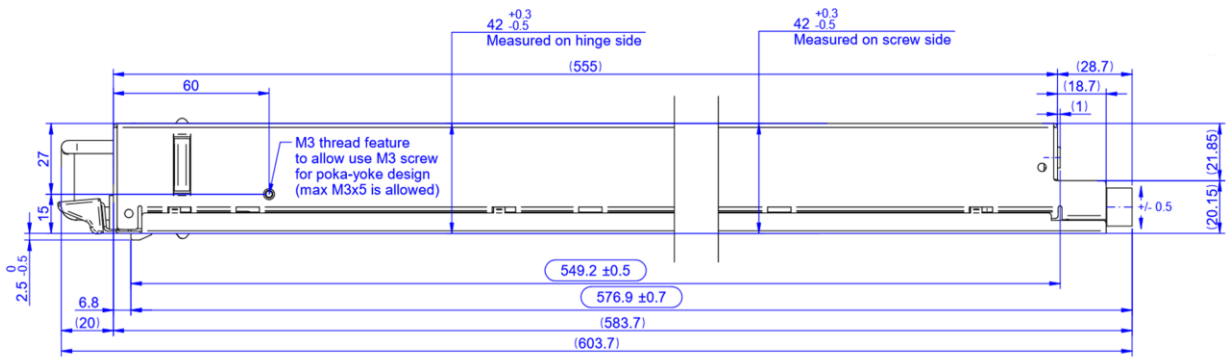


Figure 35. Side view

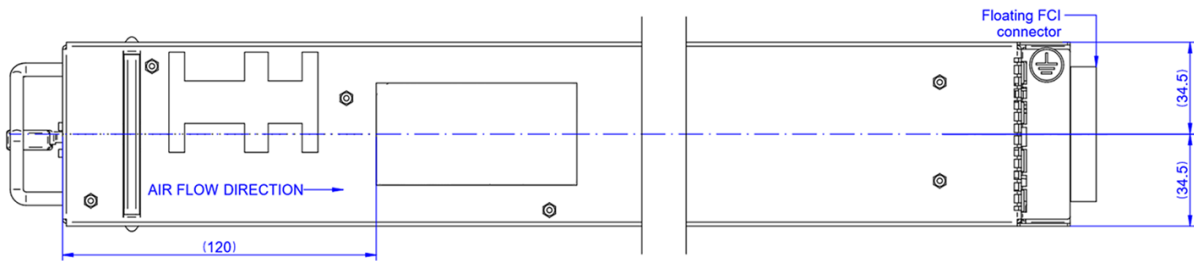


Figure 36. Top view

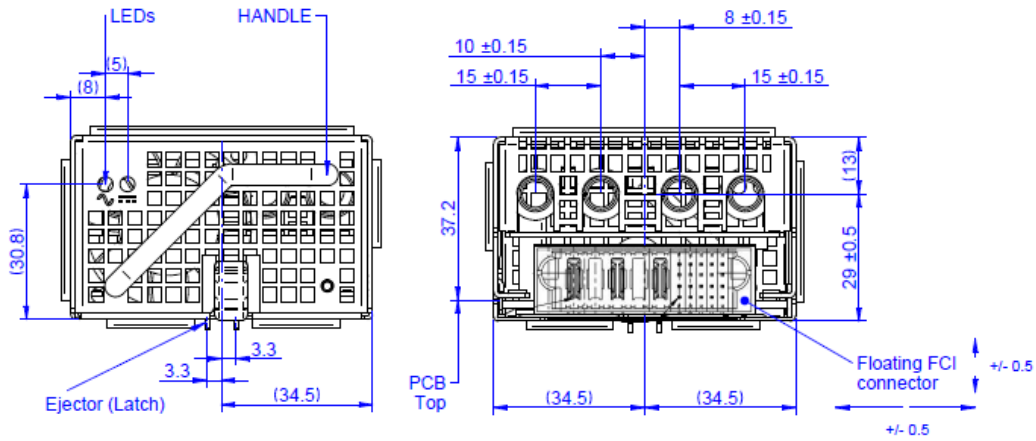


Figure 37. Front and Rear view

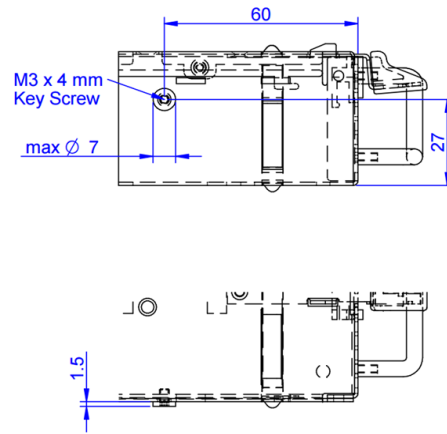


Figure 38. PFE3600-12-069RA with Key-in screw dimension (Option code K)

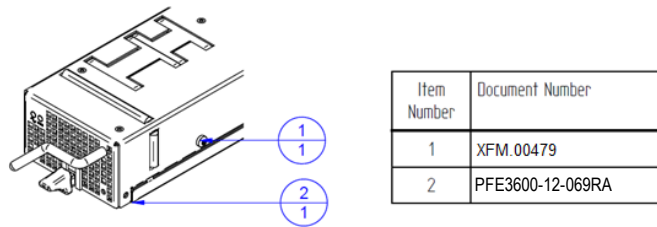
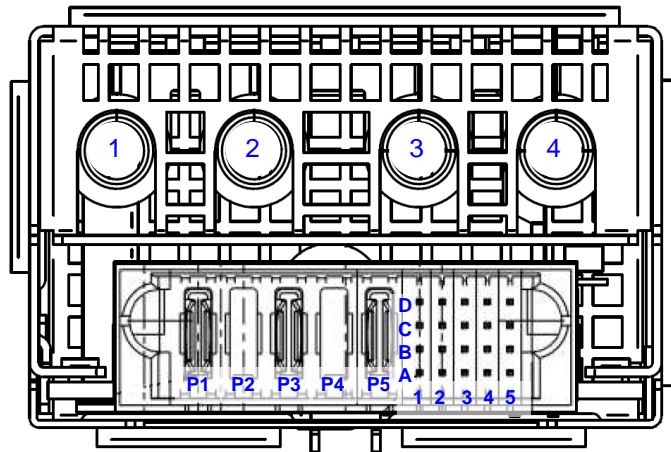


Figure 39. PFE3600-12-069RA with Key-in screw (Option code K)

14. CONNECTORS



Unit: FCI Connectors P/N 51939-768LF
 Counterpart: FCI Connectors P/N 51915-401LF
 For Main Output Pins, see section 15

Note: A1 and A2 are Trailing Pin (short pins)



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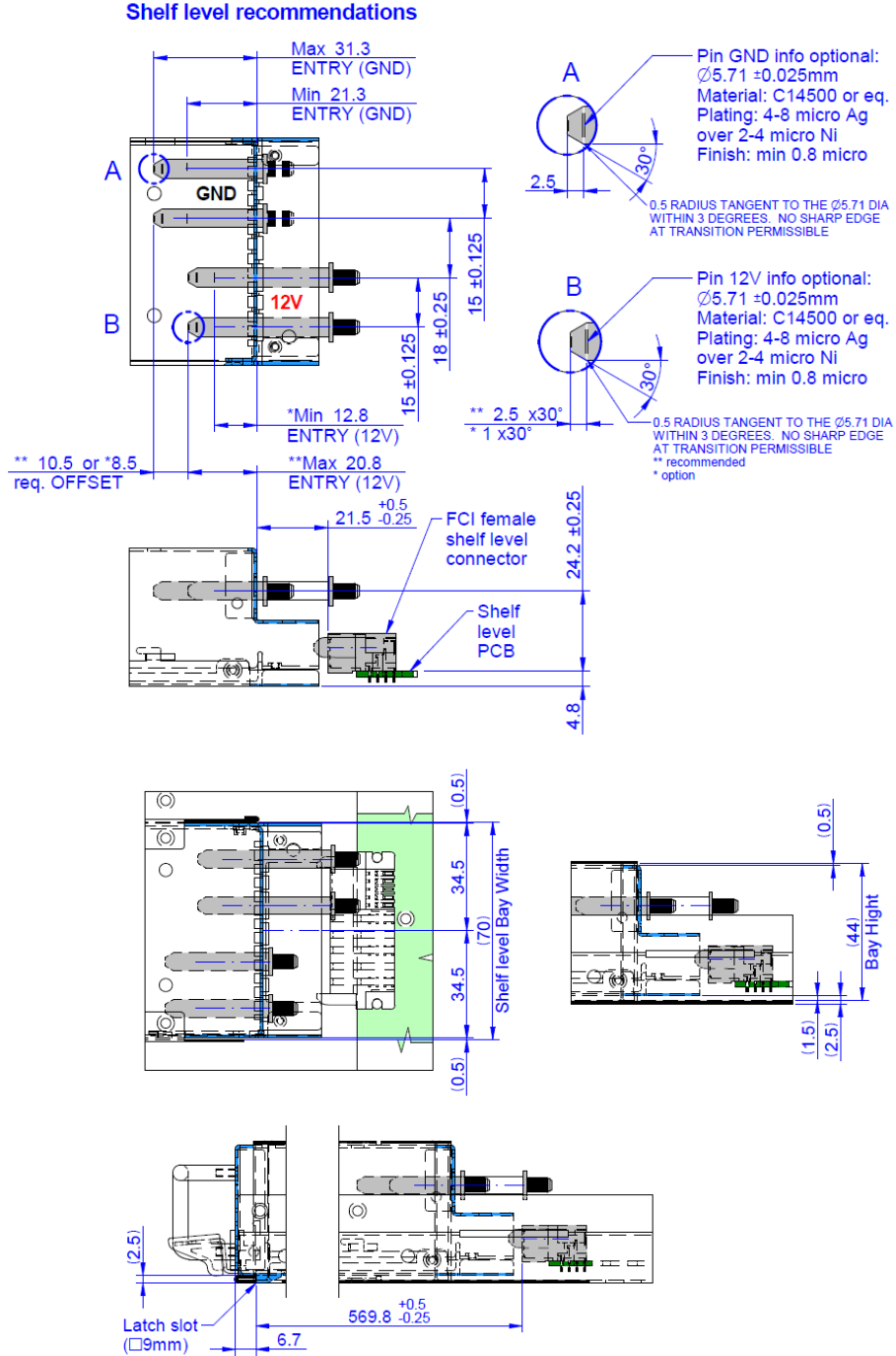
PIN	NAME	DESCRIPTION
Output		
3,4	V1	+12 VDC main output
1,2	PGND	+12 VDC main output ground
Input Pins		
P1	LIVE	AC Live Pin
P2	N.C	No metal pin connection
P3	NEUTRAL	AC Neutral Pin
P4	N.C.	No metal pin connection
P5	P.E.	Protective Earth Pin
Control Pins		
A1	PSKILL	Power supply kill (trailing pin): active-high
B1	PWOK_L	Power OK signal output: active-low
C1	INOK_L	Input OK signal: active-low
D1	PSON_L	Power supply on input: active-low
A2	PRESENT_L	Power supply present (trailing pin): active-low
B2	SGND	Signal ground* (return)
C2	SGND	Signal ground* (return)
D2	SGND	Signal ground* (return)
A3	SCL	I ² C clock signal line
B3	SDA	I ² C data signal line
C3	SMB_ALERT_L	SMB Alert signal output: active-low
D3	ISHARE	V ₁ Current share bus
A4	EEPROM_WP	EEPROM write protect
B4	RESERVED	Reserved
C4	V1_SENSE_R	Main output negative sense
D4	V1_SENSE	Main output positive sense
A5	VSB	Standby positive output
B5	VSB	Standby positive output
C5	VSB_GND	Standby Ground*
D5	VSB_GND	Standby Ground*

* These pins should be connected to PGND on the system. See Section 8 for pull up resistor settings of signal pins. All signal pins are referred to SGND



Table 9. Pin assignment

15. SHELF LEVEL CONFIGURATION (PROVISIONAL)

The recommended pin configuration below is based on company's own Shelf design and provided here as reference. Customer pin lengths within the range indicated is acceptable.



16. ACCESSORIES

ITEM	DESCRIPTION	ORDERING PN	SOURCE
	I²C Utility Windows Vista/7/8 compatible GUI to program, control and monitor PFE Front-Ends (and other I ² C units)	N/A	belfuse.com/power-solutions
	Single Connector Board Connector board to operate PFE3000-12-069RA unit. Includes an on-board USB to I ² C converter (use I ² C Utility as desktop software).	YTM.U0M00.0	Bel Power Solutions
	Key-in Screw Screw for PSU Orientation.	XFM.00479	Focus Metal

17. REVISION HISTORY

REV	DESCRIPTION	PSU PRODUCT VERSION	DATE	AUTHOR
001	Initial Release of Datasheet.	V001	09-27-2017	JSC
003	Power Management Bus reference updated; a disclaimer added to the first page. Figure 28. I ² C / SMBus Timing updated	V002 V003 V004	01-23-2018	VS
004	Change Ordering Information with Key-in screw from S366 to K. Update Power Derating and +12V OCP at 55°C Ambient Temp. Update +12VSB total regulation. Removed Pending on Safety Agency Approvals	V005	08-15-2018	GS

For more informaton on these products consult: tech.support@psbel.com

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