## BIPOLAR ANALOG INTEGRATED CIRCUITS $\mu$ PC2757TB, $\mu$ PC2758TB

## SILICON MMIC 1st FREQUENCY DOWN-CONVERTER FOR CELLULAR/CORDLESS TELEPHONE

## DESCRIPTION

The $\mu$ PC2757TB and $\mu$ PC2758TB are silicon monolithic integrated circuit designed as 1st frequency downconverter for cellular/cordless telephone receiver stage. The ICs consist of mixer and local amplifier. The $\mu$ PC2757TB features low current consumption and the $\mu \mathrm{PC} 2758$ TB features improved intermodulation. From these two version, you can chose either IC corresponding to your system design. These TB suffix ICs which are smaller package than conventional T suffix ICs contribute to reduce your system size.

The $\mu \mathrm{PC} 2757 \mathrm{~TB}$ and $\mu \mathrm{PC} 2758 \mathrm{~TB}$ are manufactured using Renesas 20 GHz ft NESAT ${ }^{\text {TMIIII }}$ silicon bipolar process. This process uses silicon nitride passivation film and gold electrodes. These materials can protect chip surface from external pollution and prevent corrosion/migration. Thus, this IC has excellent performance, uniformity and reliability.

## FEATURES

- Wideband operation
- High-density surface mounting
- Low current consumption
- Supply voltage
- Minimized carrier leakage
- Equable output impedance
: $\mathrm{fRFin}^{2}=0.1$ to 2.0 GHz , fifout $=20$ to 300 MHz
: 6-pin super minimold package
: Icc = 5.6 mA TYP. @ $\mu$ PC2757TB
Icc = 11 mA TYP. @ $\mu \mathrm{PC} 2758 \mathrm{~TB}$
: Vcc=2.7 to 3.3 V
: Due to double balanced mixer
: Single-end push-pull IF amplifier
- Built-in power save function


## APPLICATIONS

- Cellular/cordless telephone up to 2.0 GHz MAX. (example: GSM, PDC800M, PDC1.5G and so on): $\mu \mathrm{PC} 2758 \mathrm{~TB}$
- Cellular/cordless telephone up to 2.0 GHz MAX. (example: CT1, CT2 and so on): $\mu \mathrm{PC} 2757 \mathrm{~TB}$


## ORDERING INFORMATION

| Part Number | Package | Markings | Supplying Form | Product Type |
| :---: | :---: | :---: | :---: | :---: |
| $\mu \mathrm{PC} 2757 \mathrm{~TB}-\mathrm{E} 3$ | 6-pin super minimold | C1X | Embossed tape 8 mm wide. <br> Pin 1, 2, 3 face the tape perforation side. <br> Qty 3kpcs/reel. | Low current consumption |
| $\mu \mathrm{PC} 2758 \mathrm{~TB}-\mathrm{E} 3$ |  | C1Y |  | High $\mathrm{OlP}_{3}$ |

Remark To order evaluation samples, please contact your nearby sales office (Part number for sample order:
$\mu$ PC2757TB-A, $\mu$ PC2758TB-A).

## Caution Electro-static sensitive devices

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## 1. PIN CONNECTIONS



Example marking is for $\mu \mathrm{PC} 2757 \mathrm{~TB}$
$\mu \mathrm{PC} 2757 \mathrm{~TB}, \mu \mathrm{PC} 2758 \mathrm{~TB}$ in common

| Pin No. | Pin Name |
| :---: | :---: |
| 1 | RFinput |
| 2 | GND |
| 3 | LOinput |
| 4 | PS |
| 5 | Vcc |
| 6 | IFoutput |

2. PRODUCT LINE-UP ( $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}, \mathrm{Vcc}=\mathrm{Vps}=3.0 \mathrm{~V}, \mathrm{Zs}=\mathrm{ZL}=50 \Omega$ )

|  | No RF Icc (mA) | 900 MHz SSB • NF (dB) | 1.5 GHz SSB • NF (dB) | 1.9 GHz SSB • NF (dB) | 900 MHz CG (dB) | 1.5 GHz CG (dB) | 1.9 GHz CG (dB) | 900 MHz IIP3 (dBm) | $\begin{gathered} 1.5 \mathrm{GHz} \\ \mathrm{IIP} 3 \\ (\mathrm{dBm}) \end{gathered}$ | $\begin{gathered} 1.9 \mathrm{GHz} \\ \mathrm{IIP}_{3} \\ (\mathrm{dBm}) \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mu \mathrm{PC} 2757 \mathrm{~T}$ | 5.6 | 10 | 10 | 13 | 15 | 15 | 13 | -14 | -14 | -12 |
| $\mu \mathrm{PC} 2757 \mathrm{~TB}$ |  |  |  |  |  |  |  |  |  |  |
| $\mu \mathrm{PC} 2758 \mathrm{~T}$ | 11 | 9 | 10 | 13 | 19 | 18 | 17 | -13 | -12 | -11 |
| $\mu \mathrm{PC} 2758 \mathrm{~TB}$ |  |  |  |  |  |  |  |  |  |  |
| $\mu \mathrm{PC} 8112 \mathrm{~T}$ | 8.5 | 9 | 11 | 11 |  | 13 | 13 | -10 | -9 | -7 |
| $\mu \mathrm{PC8112TB}$ |  |  |  |  |  |  |  |  |  |  |


| Items <br> Part No. | 900 MHz <br> Po (sat) (dBm) | 1.5 GHz <br> Po (sat) <br> (dBm) | 1.9 GHz Po (sat) (dBm) | 900 MHz RFLo (dB) | 1.5 GHz RFıo (dB) | 1.9 GHz RFıo (dB) | IF Output <br> Configuration | Packages |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mu \mathrm{PC} 2757 \mathrm{~T}$ | -3 |  | -8 | - | - | - | Emitter follower | 6-pin minimold |
| $\mu \mathrm{PC} 2757 \mathrm{~TB}$ |  |  |  |  |  |  |  | 6-pin super minimold |
| $\mu \mathrm{PC} 2758 \mathrm{~T}$ | +1 | - | -4 | - | - | - |  | 6-pin minimold |
| $\mu \mathrm{PC} 2758 \mathrm{~TB}$ |  |  |  |  |  |  |  | 6-pin super minimold |
| $\mu \mathrm{PC} 8112 \mathrm{~T}$ | -2.5 | -3 | -3 | -80 | -57 | -55 | Open collector | 6-pin minimold |
| $\mu \mathrm{PC} 8112 \mathrm{~TB}$ |  |  |  |  |  |  |  | 6-pin super minimold |

Remark Typical performance. Please refer to ELECTRICAL CHARACTERISTICS in detail.

Cautions 1. The $\mu \mathrm{PC} 2757$ and $\mu \mathrm{PC} 2758$ 's $\mathrm{IIP}_{3}$ are calculated with $\Delta \mathrm{IM}_{3}=3$ which is the same $\mathrm{IM}_{3}$ inclination as $\mu \mathrm{PC} 8112$. On the other hand, OIP 3 of Standard characteristics in page 7 is cross point IP.
2. This document is to be specified for $\mu \mathrm{PC} 2757 \mathrm{~TB}, \mu \mathrm{PC} 2758 \mathrm{~TB}$. The other part number mentioned in this document should be referred to the data sheet of each part number.
3. INTERNAL BLOCK DIAGRAM ( $\mu$ PC2757TB, $\mu$ PC2758TB in common)


## 4. SYSTEM APPLICATION EXAMPLE

digital cellular telephone


## 5. PIN EXPLANATION (Both $\mu$ PC2757TB, 2758TB)

| Pin <br> No. | Pin <br> Name | Applied <br> Voltage (V) | Pin Voltage <br> $(\mathrm{V})^{\text {Note }}$ | Function and Application | Internal Equivalent Circuit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | RFinput | - | 1.2 | This pin is RF input for mixer designed as double balance type. This circuit contributes to suppress spurious signal with minimum LO and bias power consumption. Also this symmetrical circuit can keep specified performance insensitive to process-condition distribution. |  |
| 2 | GND | GND | - | This pin is ground of IC. <br> Must be connected to the system ground with minimum inductance. Ground pattern on the board should be formed as wide as possible. (Track length should be kept as short as possible.) | - |
| 3 | LOinput | - | $1.3$ | This pin is LO input for local buffer designed as differential amplifier. Recommendable input level is -15 to 0 dBm . Also this symmetrical circuit can keep specified performance insensitive to processcondition distribution. |  |
| 4 | PS | Vcc or GND |  | This pin is for power-save function. This pin can control ON/OFF operation with bias as follows; <br> Rise time/fall time using this pin are approximately $10 \mu \mathrm{~s}$. |  |
| 5 | Vcc | 2.7 to 3.3 | - | Supply voltage $3.0 \pm 0.3 \mathrm{~V}$ for operation. Must be connected bypass capacitor. (example: 1000 pF) to minimize ground impedance. | - |
| $6$ | IFoutput | - | 1.7 | This pin is output from IF buffer amplifier designed as single-ended push-pull type. This pin is assigned for emitter follower output with lowimpedance. In the case of connecting to high-impedance stage, please attach external matching circuit. |  |

Note Each pin voltage is measured at $\mathrm{Vcc}=3.0 \mathrm{~V}$

## 6. ABSOLUTE MAXIMUM RATINGS

| Parameter | Symbol | Conditions | Ratings | Unit |
| :--- | :---: | :--- | :---: | :---: |
| Supply Voltage | Vcc | $\mathrm{TA}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ | 5.5 | V |
| Power Dissipation of Package Allowance | PD | Mounted on $50 \times 50 \times 1.6 \mathrm{~mm}$ <br> double sided copper clad epoxy <br> glass board at $\mathrm{TA}_{\mathrm{A}}=+85^{\circ} \mathrm{C}$ | 270 | mW |
| Operating Ambient Temperature | $\mathrm{TA}_{\mathrm{A}}$ |  | -40 to +85 | ${ }^{\circ} \mathrm{C}$ |
| Storage Temperature | Tstg |  | -55 to +150 | ${ }^{\circ} \mathrm{C}$ |
| PS Pin Voltage | VPS | $\mathrm{TA}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ | 5.5 | V |

## 7. RECOMMENDED OPERATING RANGE

| Parameter | Symbol | MIN. | TYP. | MAX. | Unit |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Supply Voltage | $\mathrm{Vcc}_{\text {cc }}$ | 2.7 | 3.0 | 3.3 | V |
| Operating Ambient Temperature | $\mathrm{T}_{\mathrm{A}}$ | -40 | +25 | +85 | ${ }^{\circ} \mathrm{C}$ |
| LO Input Power | PLoin | -15 | -10 | 0 | dBm |

8. ELECTRICAL CHARACTERISTICS $\left(T_{A}=+25^{\circ} \mathrm{C}, \mathrm{Vcc}=\mathrm{VPS}_{\mathrm{F}}=3.0 \mathrm{~V}\right.$, PLoin $=\mathbf{- 1 0} \mathrm{dBm}$, $\mathrm{Zs}=\mathrm{ZL}=50 \Omega$ )

| Parameter | Symbol | Conditions | $\mu$ PC2757TB |  |  | $\mu \mathrm{PC} 2758 \mathrm{~TB}$ |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN. | TYP. | MAX. | MIN. | TYP. | MAX. |  |
| Circuit Current | Icc | No input signal | 3.7 | 5.6 | 7.7 | 6.6 | 11 | 14.8 | mA |
| RF Input Frequency | $\mathrm{frFin}^{\text {f }}$ | $\begin{aligned} & C G \geq(C G 1-3 \mathrm{~dB}) \\ & \text { fifout }=130 \mathrm{MHz} \text { constant } \end{aligned}$ | 0.1 | - | 2.0 | 0.1 | - | 2.0 | GHz |
| IF Output Frequency | $\mathrm{fiFout}^{\text {f }}$ | $\begin{aligned} & \mathrm{CG} \geq(\mathrm{CG} 1-3 \mathrm{~dB}) \\ & \mathrm{f}_{\mathrm{RF} \text { in }}=0.8 \mathrm{GHz} \text { constant } \end{aligned}$ | 20 | - | 300 | 20 | - | 300 | MHz |
| Conversion Gain 1 | CG1 | $\begin{aligned} & f_{\text {RFin }}=0.8 \mathrm{GHz}, \mathrm{fifout}=130 \mathrm{MHz} \\ & \mathrm{P}_{\text {RFin }}=-40 \mathrm{dBm}, \text { Upper local } \end{aligned}$ | 12 | 15 | 18 | 16 | 19 | 22 | dB |
| Conversion Gain 2 | CG2 | $\begin{aligned} & f_{\text {RFin }}=2.0 \mathrm{GHz}, \mathrm{fiFout}=250 \mathrm{MHz} \\ & \mathrm{P}_{\text {RFin }}=-40 \mathrm{dBm}, \text { Lower local } \end{aligned}$ | 10 | 13 | 16 | 14 | 17 | 20 | dB |
| SSB Noise Figure 1 | SSB • NF1 | $f_{\text {RFin }}=0.8 \mathrm{GHz}, \mathrm{fifout}=130 \mathrm{MHz} \text {, }$ <br> SSB mode, Upper local | - | 10 | 13 | - | 9 | 12 | dB |
| SSB Noise Figure 2 | SSB • NF2 | $f_{\text {RFin }}=2.0 \mathrm{GHz}, \mathrm{fifout}=250 \mathrm{MHz} \text {, }$ <br> SSB mode, Lower local | - | 13 | 16 | - | 13 | 15 | dB |
| Saturated Output Power 1 | Po(sat) 1 | $\begin{aligned} & f_{\text {RFin }}=0.8 \mathrm{GHz}, \mathrm{fifout}=130 \mathrm{MHz} \\ & \mathrm{P}_{\text {RFin }}=-10 \mathrm{dBm}, \text { Upper local } \end{aligned}$ | -11 | -3 | - | -7 | +1 | - | dBm |
| Saturated Output Power 2 | Po (sat) 2 | $\begin{aligned} & f_{\text {RFin }}=2.0 \mathrm{GHz}, \text { fifout }=250 \mathrm{MHz} \\ & \mathrm{P}_{\text {RFin }}=-10 \mathrm{dBm}, \text { Lower local } \end{aligned}$ | -11 | -8 | - | -7 | -4 | - | dBm |

## 9. STANDARD CHARACTERISTICS FOR REFERENCE

(Unless otherwise specified: $\mathrm{TA}_{\mathrm{A}}=+25^{\circ} \mathrm{C}, \mathrm{Vcc}=\mathrm{Vps}=3.0 \mathrm{~V}$, PLOin $=\mathbf{- 1 0 ~ d B m , ~} \mathrm{Zs}_{\mathrm{s}}=\mathrm{Z}_{\mathrm{L}}=50 \Omega$ )

| Parameter | Symbol | Conditions | Reference Value |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | $\mu \mathrm{PC} 2757 \mathrm{~TB}$ | $\mu \mathrm{PC} 2758 \mathrm{~TB}$ |  |
| 3rd Order Distortion Output Intercept Point | $\mathrm{OIP}_{3}$ | $\mathrm{f}_{\text {RFin }}=0.8$ to 2.0 GHz , $\mathrm{fifout}=0.1 \mathrm{GHz}$, Cross point IP | +5 | +11 | dBm |
| LO Leakage at RF pin | LOrf | floin $=0.8$ to 2.0 GHz | -35 | -30 | dBm |
| LO Leakage at IF pin | LOif | $\mathrm{fLoin}^{\prime}=0.8$ to 2.0 GHz | -23 | -15 | dBm |
| Circuit Current at Power Save Mode | ICC(PS) | V PS $=0.5 \mathrm{~V}$ | 0.1 | 0.1 | $\mu \mathrm{A}$ |

## 10. TEST CIRCUIT

$\mu$ PC2757TB, $\mu$ PC2758TB


* 11. ILLUSTRATION OF THE TEST CIRCUIT ASSEMBLED ON EVALUATION BOARD


Component List

| No. | Value |
| :---: | :---: |
| $\mathrm{C}_{1}, \mathrm{C}_{2}$ | 1000 pF |
| $\mathrm{C}_{3}$ to $\mathrm{C}_{5}$ | 3300 pF |

Notes 1. $35 \times 42 \times 0.4 \mathrm{~mm}$ double sided copper clad polyimide board.
2. Back side: GND pattern
3. Solder plated on pattern
4. ${ }^{\circ} \mathrm{O}$ : Through holes

## Application explanation

This IC is guaranteed on the test circuit constructed with $50 \Omega$ equipment and transmission line.
This IC, however, does not have $50 \Omega$ input/output impedance, but electrical characteristics such as conversion gain and intermodulation distortion are described herein on these conditions without impedance matching. So, you should understand that conversion gain and intermodulation distortion at input level will vary when you improve VS of RF input with external circuit ( $50 \Omega$ termination or impedance matching.)
12. TYPICAL CHARACTERISTICS ( $\mathrm{T}_{\mathrm{A}}=\mathbf{+ 2 5 ^ { \circ }} \mathbf{C}$, on Measurement Circuit)

## $12.1 \mu$ PC2757TB



CONVERSION GAIN vs. RF INPUT FREQUENCY


SSB NOISE FIGURE vs. RF INPUT FREQUENCY


CONVERSION GAIN vs. IF OUTPUT FREQUENCY


CONVERSION GAIN vs. LO INPUT POWER
CONVERSION GAIN vs. LO INPUT POWER




Remark The graphs indicate nominal characteristics.

## $12.2 \mu \mathrm{PC} 2758 \mathrm{~TB}$



SSB NOISE FIGURE vs. RF INPUT FREQUENCY


CONVERSION GAIN vs. LO INPUT POWER


CONVERSION GAIN vs. RF INPUT FREQUENCY


CONVERSION GAIN vs. IF OUTPUT FREQUENCY


CONVERSION GAIN vs. LO INPUT POWER



Remark The graphs indicate nominal characteristics.

## 13. S-PARAMETERS

## $13.1 \mu$ PC2757TB

Calibrated on pin of DUT

$\begin{array}{ll}S_{11} \\ R E F & \text { 1.0 Units }\end{array}$
$\begin{array}{lll}\stackrel{1}{\nabla} & \text { 200.0 mUnits/ } \\ h p & 104.03 \Omega & -413.42 \Omega\end{array}$ MARKER
500.0 MAZ

RF PORT
V cc $=3.0 \mathrm{~V} \quad \mathrm{~V}_{\mathrm{Ps}}=\mathrm{GND}$
$1: 500 \mathrm{MHz} \quad 104.03 \Omega-j 413.42 \Omega \quad$ START 0.050000000 GHz $2: 900 \mathrm{MHz} \quad 74.82 \Omega-\mathrm{j} 243.06 \Omega$
3:1 $500 \mathrm{MHz} 59.266 \Omega-\mathrm{j} 154.98 \Omega$
4:1 900 MHz $51.227 \Omega-\mathrm{j} 124.55 \Omega$
5:2 $500 \mathrm{MHz} 43.996 \Omega$-j95.117 $\Omega$
$\mathrm{S}_{11}$
REF
1
7
$h p$

LO PORT
$\mathrm{Vcc}=3.0 \mathrm{~V} \quad \mathrm{~V}$ Ps $=\mathrm{GND}$
$1: 500 \mathrm{MHz} \quad 114.16 \Omega$-j400.03 $\Omega \quad$ START 0.050000000 GHz 2:900 MHz $75.133 \Omega-\mathrm{j} 242.73 \Omega$ sTOP 3.000000000 GHz
3:1 $500 \mathrm{MHz} \quad 53.516 \Omega-\mathrm{j} 154.21 \Omega$
4:1 $900 \mathrm{MHz} \quad 44.789 \Omega-\mathrm{j} 124.74 \Omega$
5:2 $500 \mathrm{MHz} \quad 37.004 \Omega$-j93.828 $\Omega$
$\mathrm{S}_{22}$
Z
REF 1.0 Units
$1 \quad 200.0$ mUnits/
hp $066.38 \Omega$


## $13.2 \mu \mathrm{PC} 2758 \mathrm{~TB}$

Calibrated on pin of DUT



$\mathrm{S}_{11}$ Z
REF 1.0 Units
$\begin{array}{ll}1 & \text { 200.0 mUnits/ } \\ \nabla & 107.13 \Omega \quad-395\end{array}$
hp $107.13 \Omega \quad-395.56 \Omega$
$\mathrm{CC}=3.0 \mathrm{~V} \quad \mathrm{VPS}=\mathrm{GND}$
$\begin{array}{ll}\text { VCC }=3.0 \mathrm{~V} & \text { VPS }=\text { GND } \\ 1: 500 \mathrm{MHz} & 107.13 \Omega-\mathrm{j} 395.56 \Omega \\ \text { START } 0.050000000 \mathrm{GHz}\end{array}$ $2: 900 \mathrm{MHz} 78.711 \Omega-j 234.41 \Omega \quad$ STOP $\quad 3.000000000 \mathrm{GHz}$
3:1 $500 \mathrm{MHz} \quad 61.922 \Omega-j 148.82 \Omega$
4:1 $900 \mathrm{MHz} 52.629 \Omega-j 119.55 \Omega$
$5: 2500 \mathrm{MHz} \quad 44.766 \Omega-\mathrm{j} 90.578 \Omega$
$S_{11} \quad Z$
REF 1.0 Units
$\begin{array}{ll}1 & 200.0 \mathrm{mUnits} / \\ \nabla & 100.31 \Omega-37\end{array}$


LO PORT
$\mathrm{Vcc}=3.0 \mathrm{~V} \quad \mathrm{Vps}=\mathrm{GND}$
$\begin{array}{lll}1: 500 \mathrm{MHz} & 100.31 \Omega-\mathrm{j} 374.75 \Omega & \text { START } 0.050000000 \mathrm{GHz} \\ 2.900 \mathrm{MHz} & 73.148 \Omega-\mathrm{j} 23.07 \Omega & \text { STOP } \\ 3.000000000 \mathrm{GHz}\end{array}$
$2: 900 \mathrm{MHz} 73.148 \Omega-j 223.07 \Omega$
3:1 500 MHz $57.719 \Omega-j 144.02 \Omega$
4:1 $900 \mathrm{MHz} \quad 50.738 \Omega-j 119.52 \Omega$
5:2 $500 \mathrm{MHz} \quad 41.836 \Omega-j 90.25 \Omega$
$\mathrm{S}_{22}$
$\mathrm{S}_{22}$ Z
$\begin{array}{ll}\text { REF } & \text { 1.0 Units } \\ 1 & 200.0 \mathrm{mUnits} /\end{array}$


## 14. PACKAGE DIMENSIONS

6-PIN SUPER MINIMOLD (UNIT: mm)


## 15. NOTE ON CORRECT USE

(1) Observe precautions for handling because of electrostatic sensitive devices.
(2) Form a ground pattern as widely as possible to minimize ground impedance (to prevent undesired oscillation). Keep the track length of the ground pins as short as possible.
(3) Connect a bypass capacitor (example: 1000 pF ) to the Vcc pin.
(4) The DC cut capacitor must be attached to input pin.

## 16. RECOMMENDED SOLDERING CONDITIONS

This product should be soldered under the following recommended conditions.

| Soldering Method | Soldering Condition | Recommended Condition Symbol |
| :--- | :--- | :---: |
| Infrared Reflow | Package peak temperature: $235^{\circ} \mathrm{C}$ or below <br> Time: 30 seconds or less (at $210^{\circ} \mathrm{C}$ ) <br> Count: 3, Exposure limit: None ${ }^{\text {Note }}$ | IR35-00-3 |
| VPS | Package peak temperature: $215^{\circ} \mathrm{C}$ or below <br> Time: 40 seconds or less (at $200^{\circ} \mathrm{C}$ ) <br> Count: 3, Exposure limit: None ${ }^{\text {Note }}$ | VP15-00-3 |
| Wave Soldering | Soldering bath temperature: $260^{\circ} \mathrm{C}$ or below <br> Time: 10 seconds or less <br> Count: 1, Exposure limit: None ${ }^{\text {Note }}$ | WS60-00-1 |
| Partial Heating | Pin temperature: $300^{\circ} \mathrm{C}$ <br> Time: 3 seconds or less (per side of device) <br> Exposure limit: None ${ }^{\text {Note }}$ | - |

Note After opening the dry pack, keep it in a place below $25^{\circ} \mathrm{C}$ and $65 \%$ RH for the allowable storage period.

## Caution Do not use different soldering methods together (except for partial heating).

For details of recommended soldering conditions for surface mounting, refer to information document SEMICONDUCTOR DEVICE MOUNTING TECHNOLOGY MANUAL (C10535E).

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