

# Low Power, Class G Stereo Headphone Driver

### **DESCRIPTION**

The WM9010 is a low power stereo headphone driver designed for mobile handset and portable media player (PMP) applications. Class G amplifier technology is used to achieve high power efficiency and low quiescent current.

Stereo analogue inputs accept 1Vrms line level inputs; the Hi-Fi output drivers deliver up to 28mW into a  $32\Omega$  load.

An integrated charge pump circuit generates split-rail voltage supplies to power the ground-referenced headphone driver.

The WM9010 incorporates Wolfson Silent Switch™ technology to provide pop and click suppression whenever the WM9010 is enabled or disabled.

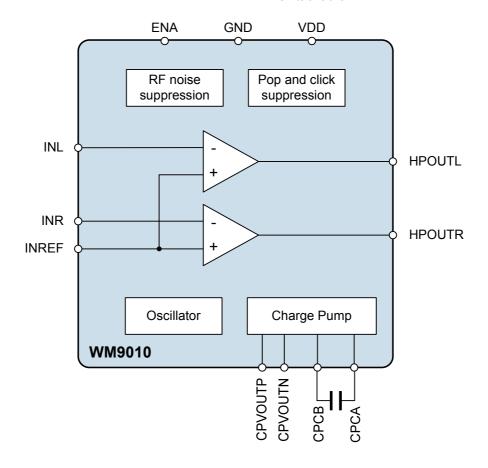
The WM9010 is supplied in a 12-pin CSP package.

#### **FEATURES**

- Hi-fi audio headphone drivers (104dB SNR 'A' weighted)
- Stereo analogue audio inputs
- · Ground-referenced, cap-less headphone outputs
- Integrated charge pump and oscillator circuits
- Low quiescent current
- RF noise suppression
- Pop and click suppression
- 12-pin CSP package (1.84 x 1.34 x 0.7mm)

#### **APPLICATIONS**

- Mobile Handsets
- Portable Media Players (PMP)
- Notebooks / Laptop computers
- LCD televisions



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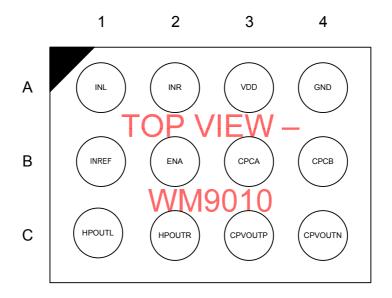
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# **PIN CONFIGURATION**

The WM9010 is supplied in a 12-pin CSP format. The pin configuration is illustrated below, showing the top-down view from above the chip.



# **ORDERING INFORMATION**

ORDER CODE	TEMPERATURE RANGE	PACKAGE	MOISTURE SENSITIVITY LEVEL	PEAK SOLDERING TEMPERATURE
WM9010ECSN/R	-40°C to +85°C	12-pin CSP (Pb-free, tape and reel)	MSL1	260°C

Note:

Reel quantity = 5000

# **PIN DESCRIPTION**

PIN NO	NAME	TYPE	DESCRIPTION
A1	INL	Analogue Input	Left channel analogue input
A2	INR	Analogue Input	Right channel analogue input
A3	VDD	Supply	Positive supply
A4	GND	Supply	Ground
B1	INREF	Analogue Input	Input reference for INL and INR
B2	ENA	Digital Input	Device Enable / Mute control
В3	CPCA	Analogue Output	Charge pump fly-back capacitor pin
B4	CPCB	Analogue Output	Charge pump fly-back capacitor pin
C1	HPOUTL	Analogue Output	Left headphone output
C2	HPOUTR	Analogue Output	Right headphone output
C3	CPVOUTP	Analogue Output	Charge pump positive supply decoupling pin
C4	CPVOUTN	Analogue Output	Charge pump negative supply decoupling pin



### **ABSOLUTE MAXIMUM RATINGS**

Absolute Maximum Ratings are stress ratings only. Permanent damage to the device may be caused by continuously operating at or beyond these limits. Device functional operating limits and guaranteed performance specifications are given under Electrical Characteristics at the test conditions specified.



ESD Sensitive Device. This device is manufactured on a CMOS process. It is therefore generically susceptible to damage from excessive static voltages. Proper ESD precautions must be taken during handling and storage of this device.

Wolfson tests its package types according to IPC/JEDEC J-STD-020B for Moisture Sensitivity to determine acceptable storage conditions prior to surface mount assembly. These levels are:

MSL1 = unlimited floor life at <30°C / 85% Relative Humidity. Not normally stored in moisture barrier bag.

MSL2 = out of bag storage for 1 year at <30°C / 60% Relative Humidity. Supplied in moisture barrier bag.

 $MSL3 = out of bag storage for 168 hours at < 30 ^{\circ}C / 60\% \ Relative Humidity. Supplied in moisture barrier bag.$ 

The Moisture Sensitivity Level for each package type is specified in Ordering Information.

CONDITION	MIN	MAX
Supply voltage (VDD)	-0.3V	2.5V
Voltage range digital input (ENA)	-0.7V	3.3V
Voltage range analogue inputs	-0.7V	VDD +0.7V
Operating temperature range, T <sub>A</sub>	-40°C	+85°C
Junction temperature, T <sub>JMAX</sub>	-40°C	+150°C
Storage temperature after soldering	-65°C	+150°C

# RECOMMENDED OPERATING CONDITIONS

PARAMETER	SYMBOL	MIN	TYP	MAX	UNIT
Supply Voltage	VDD	1.71	1.8	2.0	V
Ground	GND		0		V



# **ELECTRICAL CHARACTERISTICS**

#### **Test Conditions**

VDD = 1.8V, INREF = GND = 0V, Load resistance =  $16\Omega$ ,  $T_A$  =  $+25^{\circ}C$ , 1kHz signal unless otherwise stated.

)	Single-ended input		VDD/1.8	1		
	Single-ended input		VDD/1.8			
			-		Vrms	
		15	20	25	kΩ	
			3		pF	
IPOUTL, HPO	UTR)					
Po	$R_L = 32\Omega$ ; THD+N = 0.1%		27		mW	
	$R_L = 32\Omega$ ; THD+N = 1%		28			
	$R_L = 16\Omega$ ; THD+N = 0.1%		32			
	$R_L = 16\Omega$ ; THD+N = 1%		34			
SNR	A-weighted (20Hz – 20kHz)	95	104		dB	
THD+N	$R_L = 32\Omega; P_O = 20mW$ (20Hz – 20kHz)		-90		dB	
	$R_L = 32\Omega; P_O = 5mW$ (20Hz – 20kHz)		-91			
	$R_L = 16\Omega; P_O = 20mW$ (20Hz – 20kHz)		-91	-81		
	$R_L = 16\Omega$ ; $P_O = 5$ mW (20Hz $- 2$ 0kHz)		-87			
			0.5	2	mV	
	$R_L = 16\Omega$	80	87		dB	
	ENA = 0	95	110		dB	
PSRR	100mV pk-pk @ 217Hz on VDD		75		dB	
	HPOUTL or HPOUTR			2	nF	
		0.8 x VDD			V	
				0.2 x VDD	V	
I <sub>VDD</sub>	ENA = 1, Inputs grounded		750	1000	μΑ	
I <sub>VDD</sub>	ENA = 0		0.9	2	μA	
			13		ms	
	Po SNR THD+N PSRR	$R_{L} = 32\Omega; THD+N = 1\%$ $R_{L} = 16\Omega; THD+N = 0.1\%$ $R_{L} = 16\Omega; THD+N = 1\%$ $SNR$ $A-weighted (20Hz - 20kHz)$ $R_{L} = 32\Omega; P_{O} = 20mW$ $(20Hz - 20kHz)$ $R_{L} = 32\Omega; P_{O} = 5mW$ $(20Hz - 20kHz)$ $R_{L} = 16\Omega; P_{O} = 20mW$ $(20Hz - 20kHz)$ $R_{L} = 16\Omega; P_{O} = 5mW$ $(20Hz - 20kHz)$ $R_{L} = 16\Omega; P_{O} = 5mW$ $(20Hz - 20kHz)$ $R_{L} = 16\Omega$ $ENA = 0$ $PSRR$ $100mV pk-pk @ 217Hz on VDD$ $HPOUTL or HPOUTR$ $I_{VDD}$ $ENA = 1, Inputs grounded$	$\begin{array}{c} P_O & R_L = 32\Omega; \ THD+N = 0.1\% \\ R_L = 32\Omega; \ THD+N = 1\% \\ R_L = 16\Omega; \ THD+N = 0.1\% \\ R_L = 16\Omega; \ THD+N = 1\% \\ \hline SNR & A-weighted (20Hz - 20kHz) & 95 \\ \hline THD+N & R_L = 32\Omega; \ P_O = 20mW \\ (20Hz - 20kHz) & \\ R_L = 32\Omega; \ P_O = 5mW \\ (20Hz - 20kHz) & \\ R_L = 16\Omega; \ P_O = 20mW \\ (20Hz - 20kHz) & \\ R_L = 16\Omega; \ P_O = 5mW \\ (20Hz - 20kHz) & \\ R_L = 16\Omega; \ P_O = 5mW \\ (20Hz - 20kHz) & \\ \hline R_L = 16\Omega & 80 \\ \hline ENA = 0 & 95 \\ \hline PSRR & 100mV \ pk-pk @ 217Hz \ on \ VDD \\ \hline HPOUTL \ or \ HPOUTR & \\ \hline \\ I_{VDD} & ENA = 1, \ Inputs \ grounded \\ \hline \end{array}$	$\begin{array}{c} P_O \\ R_L = 32\Omega; \ THD+N = 0.1\% \\ R_L = 32\Omega; \ THD+N = 1\% \\ R_L = 16\Omega; \ THD+N = 0.1\% \\ R_L = 16\Omega; \ THD+N = 1\% \\ SNR \\ A-weighted (20Hz - 20kHz) \\ \hline THD+N \\ R_L = 32\Omega; \ P_O = 20mW \\ (20Hz - 20kHz) \\ \hline R_L = 32\Omega; \ P_O = 5mW \\ (20Hz - 20kHz) \\ \hline R_L = 16\Omega; \ P_O = 20mW \\ (20Hz - 20kHz) \\ \hline R_L = 16\Omega; \ P_O = 5mW \\ (20Hz - 20kHz) \\ \hline R_L = 16\Omega; \ P_O = 5mW \\ (20Hz - 20kHz) \\ \hline R_L = 16\Omega \Rightarrow 0 \\ \hline SNR \\ \hline S$	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	



#### **TERMINOLOGY**

 Signal-to-Noise Ratio (dB) – SNR is the difference in level between a full scale output signal and the device output noise with no signal applied, measured over a bandwidth of 20Hz to 20kHz. (No Auto-zero or Mute function is employed).

- 2. Total Harmonic Distortion (dB) THD is the difference in level between a 1kHz reference sine wave output signal and the sum of the harmonics of the output signal. The amplitude of the fundamental frequency of the output signal is compared to the RMS value of the sum of the harmonics and expressed as a ratio.
- 3. Total Harmonic Distortion plus Noise (dB) THD+N is the difference in level between a 1kHz reference sine wave output signal and all noise and distortion products in the audio band. The amplitude of the fundamental reference frequency of the output signal is compared to the RMS value of all other noise and distortion products and expressed as a ratio.
- 4. Channel Separation (L/R) (dB) is a measure of the coupling between left and right channels. A full scale signal is applied to the left channel only, and the right channel amplitude is measured. Next, a full scale signal is applied to the right channel only, and the left channel amplitude is measured. The worst case channel separation is quoted; this is the difference in level between the full-scale output and the cross-channel output signal level, expressed as a ratio.
- 5. Power Supply Rejection Ratio (dB) PSRR is a measure of ripple attenuation between a power supply rail and a signal output path. With the signal path idle, a small sine wave ripple is applied to power supply rail. The amplitude of the supply ripple is compared to the amplitude of the output signal generated and is expressed as a ratio.
- 6. Mute attenuation This is a measure of the difference in level between the full scale output signal and the output with mute applied (ie. ENA = logic 0).
- 7. All performance measurements are carried out with 20kHz AES17 low pass filter for distortion measurements, and an A-weighted filter for noise measurement. Failure to use such a filter will result in higher THD and lower SNR and Dynamic Range readings than are found in the Electrical Characteristics. The low pass filter removes out-of-band noise; although it is not audible, it may affect dynamic specification values.



### **DEVICE DESCRIPTION**

#### INTRODUCTION

The WM9010 is a low power stereo headphone driver designed for mobile handset and portable media player (PMP) applications. It is packaged in a 12-pin CSP.

The device comprises two analogue input pins, each accepting line signals up to 1Vrms. The signal path gain is fixed at 0dB. The headphone output drivers deliver up to 28mW into a  $32\Omega$  load.

The WM9010 incorporates Class G technology to achieve high efficiency and low quiescent current. An integrated charge pump circuit is used to generate the split (positive and negative) power rails from a single VDD supply. The ground-referenced headphone driver design reduces the device power consumption and also eliminates external DC-blocking capacitors on the audio output path.

The WM9010 is enabled when a logic high level is detected on ENA. Note that the ENA pin can support digital logic levels up to 2.7V.

Wolfson's Silent Switch™ technology is incorporated in order to minimise 'pop' noise whenever the WM9010 is enabled or disabled.

A power on reset circuit ensures correct start-up and shut-down when VDD is switched on or off. The WM9010 is held in reset when the ENA pin is held low, offering a low-power standby state.

Short circuit and thermal protection is also provided.

#### **HEADPHONE DRIVER**

The WM9010 has two analogue input pins, INL and INR. The input signals are referenced to the INREF pin, which is provided as a quiet ground reference connection. The maximum analogue input signal level varies with VDD, but is typically 0dBV (1Vrms) when VDD = 1.8V. This is suitable for single-ended connection to line level input signals.

The headphone output drivers are capable of driving up to 28mW into a  $32\Omega$  load such as a stereo headset or headphones. The outputs are ground-referenced, eliminating any requirement for AC coupling capacitors. This is achieved by having separate positive and negative supply rails powered by an on-chip charge pump.

A pop-suppression circuit ensures that DC offsets are minimised, suppressing 'pop' noise and reducing power consumption. To obtain optimal DC offsets, the device should be powered on with no signal on the input pins.

It is recommended to connect a zobel network to the headphone output pins HPOUTL and HPOUTR for best audio performance in all applications. The components of the zobel network have the effect of dampening high frequency oscillations or instabilities that can arise outside the audio band under certain conditions. Possible sources of these instabilities include the inductive load of a headphone coil or an active load in the form of an external line amplifier. The capacitance of lengthy cables or PCB tracks can also lead to amplifier instability. The zobel network should comprise of a  $20\Omega$  resistor and 100nF capacitor in series with each other, as illustrated in Figure 1.

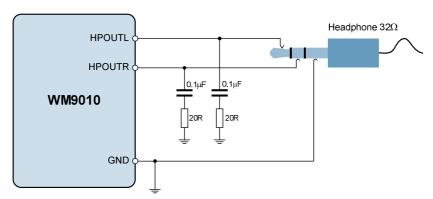


Figure 1 External Connections for HPOUTL and HPOUTR



### **CHARGE PUMP**

The WM9010 incorporates a charge pump circuit, which generates the supply rails for the headphone output drivers. The charge pump is powered from VDD, and generates split rails CPVOUTP and CPVOUTN. The circuit is adaptive according to the audio signal conditions, supporting the Class G operation and ensuring optimum circuit configuration at all times. The switching clock for the charge pump is generated internally.

The external connections for the charge pump are illustrated in Figure 2. A fly-back capacitor is connected between the CPCA and CPCB pins. De-coupling capacitors are required on CPVOUTP and CPVOUTN. An input decoupling capacitor may also be required at VDD, depending upon the system configuration.

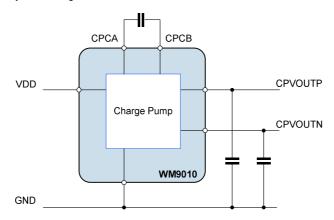


Figure 2 External Connections for Charge Pump



# **APPLICATIONS INFORMATION**

# **RECOMMENDED EXTERNAL COMPONENTS**

Figure 3 provides a summary of recommended external components for WM9010. Note that the actual requirements may differ according to the specific target application.

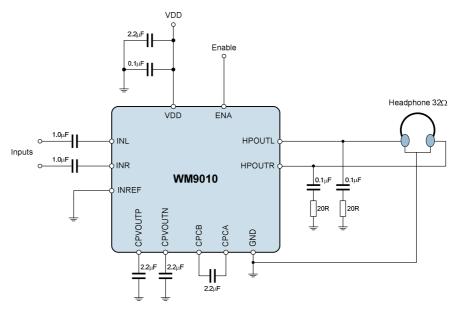
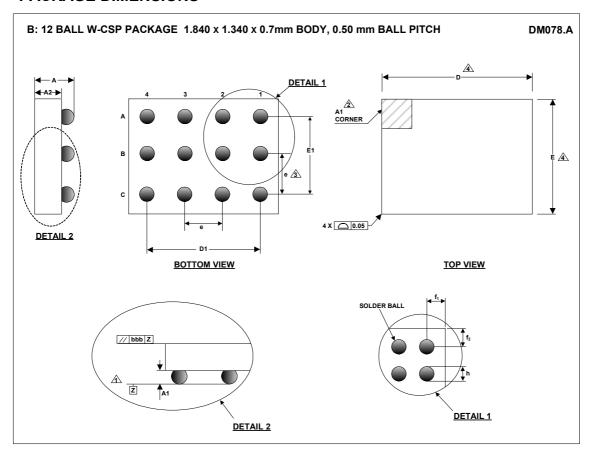


Figure 3 WM9010 Recommended External Components Diagram

# **PCB LAYOUT CONSIDERATIONS**

Poor PCB layout will degrade the performance and be a contributory factor in EMI, ground bounce and resistive voltage losses. All external components should be placed as close to the WM9010 device as possible, with current loop areas kept as small as possible.

# **PACKAGE DIMENSIONS**



Symbols	Dimensions (mm)			
	MIN	NOM	MAX	NOTE
Α	0.675	0.7	0.725	
A1	0.187	0.2	0.213	
A2	0.488	0.500	0.512	
D	1.820	1.840	1.860	
D1		1.500 BSC		
E	1.320	1.340	1.360	
E1		1.000 BSC		
е		0.500 BSC		3
f <sub>1</sub>	0.160			
f <sub>2</sub>	0.160			
h		0.267		

- NOTES:

  1. PRIMARY DATUM -Z- AND SEATING PLANE ARE DEFINED BY THE SPHERICAL CROWNS OF THE SOLDER BALLS.

  2. A1 CORNER IS IDENTIFIED BY INK/LASER MARK ON TOP PACKAGE.

  3. 've' REPRESENTS THE BASIC SOLDER BALL GRID PITCH.

  4. THIS DRAWING IS SUBJECT TO CHANGE WITHOUT NOTICE.

  5. FOLLOWS JEDEC DESIGN CUIDE MO-211-C.



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# **REVISION HISTORY**

DATE	REV	ORIGINATOR	CHANGES
18/02/11	4.0	CT/JMacD	Product status updated to Production Data
			Max and Min Electrical Characteristics Figures added.
	4.0	JMacD/IS	Bandwidth specifications added to SNR and THD+N in Electrical Characteristics



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STPA002OD-4WX NCP2823BFCT1G MAX9717DETA+T MAX9717CETA+T MAX9724AEBC+TG45 LA4450L-E IS31AP2036A
CLS2-TR MAX9723DEBE+T TDA7563ASMTR AS3561-DWLT SSM2517CBZ-R7 MP1720DH-12-LF-P SABRE9601K THAT1646W16
U MAX98396EWB+ PAM8965ZLA40-13 BD37532FV-E2 BD5638NUX-TR BD37512FS-E2 BD37543FS-E2 BD3814FV-E2

TPA3140D2PWPR TS2007EIJT IS31AP2005-DLS2-TR AS3410-EQFP-500 FDA4100LV MAX98306ETD+T TS4994EIJT

NCP2820FCT1G NCP2823AFCT2G NCS2211MNTXG CPA2233CQ16-A1 OPA1604AIPWR TDA7492 SSM2519ACBZ-R7

ZXCD1210JB16TA