

## Features

- $\varnothing 0.500 \mathrm{~mm}$ active area
- Low noise
- High gain
- Long term stability


## Description

The AD500-9-8015 TO52 is an Avalanche Photodiode Amplifier Hybrid containing a $0.196 \mathrm{~mm}^{2}$ active area APD chip integrated with an internal transimpedance amplifier. Hermetically packaged in a TO-52 with a flat borosilicate glass window cap.

Applications

- Precision photometry
- Analytical instruments
- Medical equipment
- Low light sensor

Absolute maximum ratings

| Symbol | Parameter | Min | Max | Unit |
| :---: | :--- | :---: | :---: | :---: |
| $\mathrm{T}_{\text {STG }}$ | Storage Temp | -55 | +125 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\text {Op }}$ | Operating Temp | 0 | +60 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\text {soldering }}$ | Soldering Temp | - | +240 | ${ }^{\circ} \mathrm{C}$ |
| P | Power Dissipation | - | 360 | mW |
| $\mathrm{~V}_{\mathrm{cc}}$ | Single Supply Voltage | +4.5 | +11 | V |
| $\mathrm{I}_{\mathrm{cc}}$ | Supply Current | - | 26 | mA |

Schematic


Spectral response @ M = 100


Electro-optical characteristics @ $23^{\circ} \mathrm{C}\left(V_{c \mathrm{c}}=\right.$ single supply $+5 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=50 \mathrm{~W}$ unless otherwise specified)

| Symbol | Characteristic | Test-Condition | Min | Typ | Max | Unit |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| $f_{-3 \mathrm{~dB}}$ | Frequency Response | $-3 \mathrm{~dB} @ 905 \mathrm{~nm}$ | --- | 100 | --- | MHz |
| I | Sensitivity* | $\lambda=905 \mathrm{~nm} ; \mathrm{M}=100$ | --- | 1160 | --- | $\mathrm{KV} / \mathrm{W}$ |
| $\mathrm{I}_{\mathrm{cc}}$ | Supply Current | Dark state | --- | 25 | 26 | mA |

[^0]These devices are sensitive to electrostatic discharge. Please use ESD precautions when handling.

Avalanche photodiode data @ $23^{\circ} \mathrm{C}$

| Symbol | Characteristic | Test-Condition | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ID | Dark Current | $\mathrm{M}=100$ (see note 1) | --- | 0.5 | 5.0 | nA |
| C | Capacitance | $\mathrm{M}=100$ (see note 1) | --- | 1.2 | --- | pF |
| $V_{B R}$ | Breakdown Voltage | $\mathrm{I}_{\mathrm{D}}=2 \mu \mathrm{~A}$ | 160 | 200 | --- | V |
|  | Temperature Coefficient of $\mathrm{V}_{\text {BR }}$ |  | --- | 1.55 | --- | V/K |
|  | Responsivity | $\mathrm{M}=100 ;=0 \mathrm{~V} ; \lambda=905 \mathrm{~nm}$ | 55 | 60 | --- | A/W |
| $\Delta f_{3 \mathrm{~dB}}$ | Bandwidth | -3dB | --- | 0.5 | --- | GHz |
| $\mathrm{t}_{\mathrm{r}}$ | Rise Time | $\mathrm{M}=100$ | --- | 550 | --- | ps |
|  | Optimum Gain |  | 50 | 60 | --- |  |
|  | "Excess Noise" factor | $M=100$ | --- | 2.5 | --- |  |
|  | "Excess Noise" index | $\mathrm{M}=100$ | --- | 0.2 | --- |  |
|  | Noise Current | $\mathrm{M}=100$ | --- | 1.0 | --- | $\mathrm{pA} / \mathrm{Hz}^{1 / 2}$ |
|  | Max Gain |  | 200 | --- | --- |  |
| NEP | Noise Equivalent Power | $M=100 ; \lambda=905 \mathrm{~nm}$ | --- | $2.0 \times 10^{-14}$ | --- | $\mathrm{W} / \mathrm{Hz}^{1 / 2}$ |

Note 1: Measurement conditions: Setup of photo current 1 nA at $\mathrm{M}=1$ and irradiated by a $880 \mathrm{~nm}, 80 \mathrm{~nm}$ bandwidth LED.
Increase the photo current up to $100 \mathrm{nA},(\mathrm{M}=100)$ by internal multiplication due to an increasing bias voltage.
Transimpedance amplifier data @ $25^{\circ} \mathrm{C}$
$\left(\mathrm{Vcc}=+4.5 \mathrm{~V}\right.$ to $+11 \mathrm{~V}, \mathrm{TA}=0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 50 \Omega$ load between OUT+ and OUT-. Typical values are at $\mathrm{TA}=25^{\circ} \mathrm{C}, \mathrm{Vcc}=+5 \mathrm{~V}$ )

| Parameter | Test-Condition | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Supply Voltage |  | +4.5 | +5 | +11 | V |
| Supply Current |  | --- | 25 | 26 | mA |
| Transimpedance | Differential, measured with $40 \mu \mathrm{~A}$ p-p signal | 16 | 20 | 24 | K $\Omega$ |
| Output impedance | Single ended per side | 40 | 50 | 60 | $\Omega$ |
| Maximum Differential Output Voltage | Input = 2 mA p -p with $50 \Omega$ differential termination | --- | 600 | --- | mV p-p |
| Input Referred RMS Noise | TO-5 package, see note 3 | --- | 26.5 | --- | nA |
| Input Referred Noise Density | See note 3 | --- | 3.0 | --- | $\mathrm{pA} / \mathrm{Hz}^{1 / 2}$ |
| Small signal bandwidth | Source capacitance $=1.2 \mathrm{pF}$, see note 2 | 180 | 240 | --- | MHz |
| Low Frequency Cutoff | -3 dB, input < $20 \mu \mathrm{~A}$ DC | --- | 5 | --- | KHz |
| Transimpedance Linear Range | Peak to peak 0.95 < linearity < 1.05 | $\pm 25$ | $\pm 30$ | --- | $\mu \mathrm{A} p-\mathrm{p}$ |
| Power Supply Rejection Ratio (PSRR) |  | --- | 40 | --- | dB |

Note 2: Source capacitance for AD500-9-8015-TO52 is the capacitance of APD.
Note 3: Input referred noise is calculated as RMS output noise/ (gain at $f=100 \mathrm{Mhz}$ ). Noise density is (input referred noise)/Vbandwidth.

## TRANSFER CHARACTERISTICS

The circuit used is an avalanche photodiode directly coupled to a high speed data handling transimpedance amplifier. The output of the APD (light generated current) is applied to the input of the amplifier. The amplifier output is in the form of a differential voltage pulsed signal.

The APD responsivity curve is provided in Fig. 2. The term Amps/Watt involves the area of the APD and can be expressed as Amps $/ \mathrm{mm}^{2} / \mathrm{Watts} / \mathrm{mm}^{2}$, where the numerator applies to the current generated divided by the area of the detector, the denominator refers to the power of the radiant energy present per unit area. As an example assume a radiant input of 1 microwatt at 850 nm . The APD's corresponding responsivity is $0.4 \mathrm{~A} / \mathrm{W}$.

If energy in $=1 \mu \mathrm{~W}$, then the current from the $\mathrm{APD}=(0.4 \mathrm{~A} / \mathrm{W}) \times\left(1 \times 10^{-6} \mathrm{~W}\right)=0.4 \mu \mathrm{~A}$. We can then factor in the typical gain of the APD of 100 , making the input current to the amplifier $40 \mu \mathrm{~A}$.

## APPLICATION NOTES

The AD500-9-8015-TO52 is a high speed optical data receiver. It incorporates an internal transimpedance amplifier with an avalanche photodiode.
This detector requires +4.5 V to +11 V voltage supply for the amplifier and a high voltage supply ( $100-240 \mathrm{~V}$ ) for the APD. The internal APD follows the gain curve published for the AD500-9-TO52-S1 avalanche photodiode. The transimpedance amplifier provides differential output signals in the range of 200 millivolts differential.

In order to achieve highest gain, the avalanche photodiode needs a positive bias voltage (Fig. 1). However, a current limiting resistor must be placed in series with the photodiode bias voltage to limit the current into the transimpedance amplifier. Failure to limit this current may result in permanent failure of the device. The suggested initial value for this limiting resistor is 390 KOhm.

When using this receiver, good high frequency placement and routing techniques should be followed in order to achieve maximum frequency response. This includes the use of bypass capacitors, short leads and careful attention to impedance matching. The large gain bandwidth values of this device also demand that good shielding practices be used to avoid parasitic oscillations and reduce output noise.

Fig. 1: APD gain vs bias voltage


Fig. 2: APD Spectral response ( $\mathrm{M}=1$ )


Fig. 3: Amplifier bandwith vs temperature


Fig.4: APD Capacitance vs voltage


Fig. 5: Differential gain vs. supply


Fig. 6: Amplifier gain vs. frequency


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[^0]:    * Sensitivity = APD responsivity (0.58 A/W X 100 gain) x TIA gain (20 K)

