

1MHZ CMOS Rail-to-Rail IO Opamp with RF Filter

Features

Single-Supply Operation from +1.8V ~ +6V

• Rail-to-Rail Input / Output

Gain-Bandwidth Product: 1MHz (Typ.)

Low Input Bias Current: 1pA (Typ.)

Low Offset Voltage: 3.5mV (Max.)

• Quiescent Current: 75µA per Amplifier (Typ.)

• Embedded RF Anti-EMI Filter

Operating Temperature: -40°C ~ +125°C

• Small Package:

MCP6001 Available in SOT23-5 and SC70-5 Packages
MCP6002 Available in SOP-8 and MSOP-8 Packages
MCP6004 Available in SOP-14 and TSSOP-14 Packages

General Description

The MCP6001 family have a high gain-bandwidth product of 1MHz, a slew rate of $0.8V/\,\mu\,s$, and a quiescent current of $75\,\mu$ A/amplifier at 5V. The MCP6001 family is designed to provide optimal performance in low voltage and low noise systems. They provide rail-to-rail output swing into heavy loads. The input common mode voltage range includes ground, and the maximum input offset voltage is 3.5mV for MCP6001 family. They are specified over the extended industrial temperature range (-4 $^{\circ}$ C to +125 $^{\circ}$ C). The operating range is from 1.8V to 6V. The MCP6001 single is available in Green SC70-5 and SOT23-5 packages. The MCP6002 dual is available in Green SOP-14 and TSSOP-14 packages.

Applications

- ASIC Input or Output Amplifier
- Sensor Interface
- Medical Communication
- Smoke Detectors

- Audio Output
- Piezoelectric Transducer Amplifier
- Medical Instrumentation
- · Portable Systems

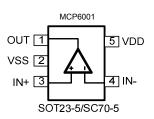
Ordering Information

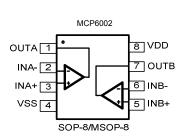
DEVICE	Package Type	MARKING	Packing	Packing Qty
MCP6001M5/TR	SOT23-5	6001/AANN ^(Note₁)	REEL	3000pcs/reel
MCP6001M7/TR	SC70-5	6001/AANN ^(Note₁)	REEL	3000pcs/reel
MCP6002M/TR	SOP-8L	MCP6002	REEL	2500pcs/reel
MCP6002MM/TR	MSOP-8L	6002	REEL	3000pcs/reel
MCP6004M/TR	SOP-14L	MCP6004	REEL	2500pcs/reel
MCP6004MT/TR	TSSOP-14L	P6004	REEL	2500pcs/reel

Note1:" NN" =year and month code.Alphanumeric traceability code.



Pin Configuration





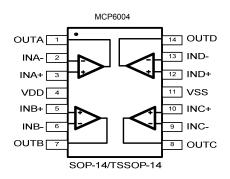


Figure 1. Pin Assignment Diagram

Absolute Maximum Ratings

Condition	Min	Max				
Power Supply Voltage (V _{DD} to Vss)	-0.5V	+7.5V				
Analog Input Voltage (IN+ or IN-)	Vss-0.5V	V _{DD} +0.5V				
PDB Input Voltage	Vss-0.5V	+7V				
Operating Temperature Range	-40°C	+125°C				
Junction Temperature	+16	0°C				
Storage Temperature Range	-55°C	+150°C				
Lead Temperature (soldering, 10sec)	+26	0°C				
Package Thermal Resistance (T _A =+25°C)	•					
SOP-8, θ _{JA}	125°	125°C/W				
MSOP-8, θ _{JA}	216°	216°C/W				
SOT23-5, θ _{JA}	190°	190°C/W				
SC70-5, θ _{JA}	333°C/W					
ESD Susceptibility						
НВМ	6KV					
MM	400V					

Note: Stress greater than those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions outside those indicated in the operational sections of this specification are not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

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Electrical Characteristics

(At VS = +5V, RL = $100k\Omega$ connected to VS/2, and VOUT = VS/2, unless otherwise noted.)

	SYMBOL	CONDITIONS	MCP6001/2/4				
PARAMETER			TYP	MIN/MAX OVER TEMPERATURE			
			+25℃	+25℃	-40℃ to +85℃	UNITS	MIN/MAX
INPUT CHARACTERISTICS							
Input Offset Voltage	Vos	V _{CM} = V _S /2	0.8	3.5	5.6	mV	MAX
Input Bias Current	I _B		1			pA	TYP
Input Offset Current	los		1			pA	TYP
Common-Mode Voltage Range	V _{CM}	V _S = 5.5V	-0.1 to +5.6			٧	TYP
Common-Mode Rejection Ratio	CMRR	$V_S = 5.5V$, $V_{CM} = -0.1V$ to 4V	70	62	62	dB	MINI
Common-Mode Rejection Ratio	CIVIRK	$V_S = 5.5V$, $V_{CM} = -0.1V$ to 5.6V	68	56	55		MIN
Open Lean Voltage Cain	^	$R_L = 5k\Omega$, $V_O = +0.1V$ to +4.9V	80	70	70	dB	MIN
Open-Loop Voltage Gain	A _{OL}	$R_L = 10k\Omega$, $V_O = +0.1V$ to +4.9V	100	94	85		IVIIIN
Input Offset Voltage Drift	$\Delta V_{OS}/\Delta_T$		2.7			μV/°C	TYP
OUTPUT CHARACTERISTICS							
	V _{OH}	R _L = 100kΩ	4.997	4.980	4.970	V	MIN
Outnut Valtage Cuing from Deil	V _{OL}	R _L = 100kΩ	5	20	30	mV	MAX
Output Voltage Swing from Rail	V _{OH}	$R_L = 10k\Omega$	4.992	4.970	4.960	V	MIN
	V _{OL}	$R_L = 10k\Omega$	8	30	40	mV	MAX
Outrout Comment	I _{SOURCE}	D = 100 to 1/ /2	84	60	45	A	MINI
Output Current	I _{SINK}	$R_L = 10\Omega$ to $V_S/2$	75	60	45	mA	MIN
POWER SUPPLY							
On another Walter and Danier				1.8	1.8	V	MIN
Operating Voltage Range				6	6	V	MAX
Power Supply Rejection Ratio	PSRR	$V_S = +2.5V \text{ to } +6V, V_{CM} = +0.5V$	82	60	58	dB	MIN
Quiescent Current / Amplifier	ΙQ		75	110	125	μΑ	MAX
DYNAMIC PERFORMANCE (CL	= 100pF)			•			
Gain-Bandwidth Product	GBP		1			MHz	TYP
Slew Rate	SR	G = +1, 2V Output Step	0.8			V/µs	TYP
Settling Time to 0.1%	t _S	G = +1, 2V Output Step	5.3			μs	TYP
Overload Recovery Time		V _{IN} ·Gain = V _S	2.6			μs	TYP
NOISE PERFORMANCE							
Voltago Noigo Dessity	_	f = 1kHz	27			nV/\sqrt{Hz}	TYP
Voltage Noise Density	e _n	f = 10kHz	20			nV/\sqrt{Hz}	TYP

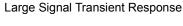


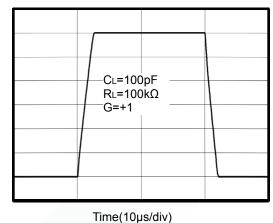
Output Voltage (250mV/div)

Typical Performance characteristics

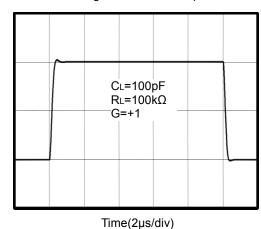
At T_A =+25°C, Vs=5V, R_L =100K Ω connected to V_S /2 and V_{OUT} = V_S /2, unless otherwise noted.

Output Voltage (50mV/div)

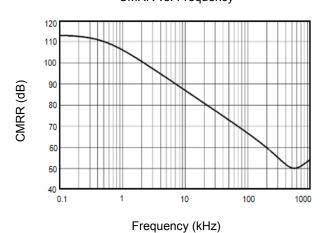




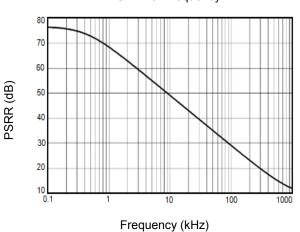
Small Signal Transient Response



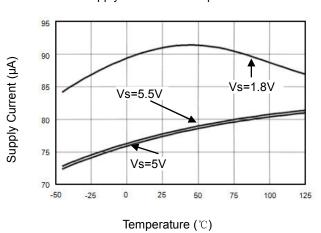
CMRR vs. Frequency



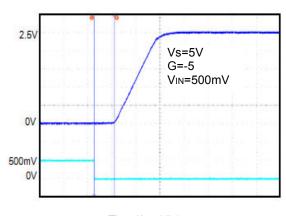
PSRR vs. Frequency



Supply Current vs. Temperature



Overload Recovery Time



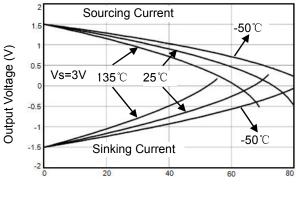
Time(2µs/div)



Typical Performance characteristics

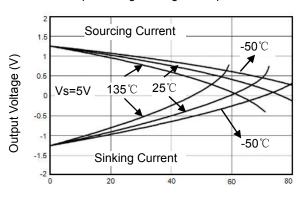
At T_A =+25°C, R_L =100K Ω connected to V_S /2 and V_{OUT} = V_S /2, unless otherwise noted.

Output Voltage Swing vs. Output Current



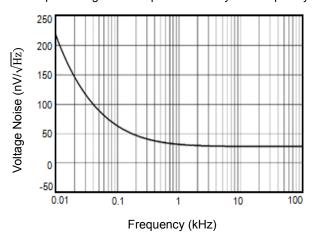
Output Current(mA)

Output Voltage Swing vs.Output Current

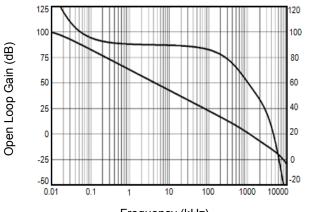


Output Current(mA)

Input Voltage Noise Spectral Density vs. Frequency



Open Loop Gain, Phase Shift vs. Frequency



Frequency (kHz)

Phase Shift (Degrees)



Application Note

Size

MCP6001 family series op amps are unity-gain stable and suitable for a wide range of general-purpose applications. The small footprints of the MCP6001 family packages save space on printed circuit boards and enable the design of smaller electronic products.

Power Supply Bypassing and Board Layout

MCP6001 family series operates from a single 1.8V to 6V supply or dual ± 0.9 V to ± 3 V supplies. For best performance, a 0.1μ F ceramic capacitor should be placed close to the V_{DD} pin in single supply operation. For dual supply operation, both V_{DD} and V_{SS} supplies should be bypassed to ground with separate 0.1μ F ceramic capacitors.

Low Supply Current

The low supply current (typical 75µA per channel) of MCP6001 family will help to maximize battery life. They are ideal for battery powered systems

Operating Voltage

MCP6001 family operates under wide input supply voltage (1.8V to 6V). In addition, all temperature specifications apply from -40 °C to +125 °C. Most behavior remains unchanged throughout the full operating voltage range. These guarantees ensure operation throughout the single Li-Ion battery lifetime

Rail-to-Rail Input

The input common-mode range of MCP6001 family extends 100mV beyond the supply rails (V_{SS} -0.1V to V_{DD} +0.1V). This is achieved by using complementary input stage. For normal operation, inputs should be limited to this range.

Rail-to-Rail Output

Rail-to-Rail output swing provides maximum possible dynamic range at the output. This is particularly important when operating in low supply voltages. The output voltage of MCP6001 family can typically swing to less than 10mV from supply rail in light resistive loads (>100k Ω), and 60mV of supply rail in moderate resistive loads (10k Ω).

Capacitive Load Tolerance

The MCP6001 family is optimized for bandwidth and speed, not for driving capacitive loads. Output capacitance will create a pole in the amplifier's feedback path, leading to excessive peaking and potential oscillation. If dealing with load capacitance is a requirement of the application, the two strategies to consider are (1) using a small resistor in series with the amplifier's output and the load capacitance and (2) reducing the bandwidth of the amplifier's feedback loop by increasing the overall noise gain. Figure 2 shows a unity gain follower using the series resistor strategy. The resistor isolates the output from the capacitance and, more importantly, creates a zero in the feedback path that compensates for the pole created by the output capacitance.

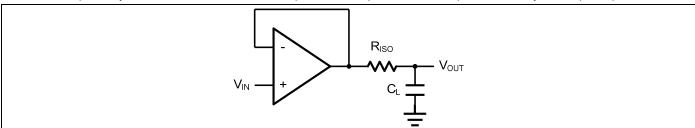


Figure 2 Indirectly Driving a Capacitive Load Using Isolation Resistor

The bigger the R_{ISO} resistor value, the more stable V_{OUT} will be. However, if there is a resistive load R_L in parallel with the capacitive load, a voltage divider (proportional to R_{ISO}/R_L) is formed, this will result in a gain error.

The circuit in Figure 3 is an improvement to the one in Figure 2. R_F provides the DC accuracy by feed-forward the V_{IN} to R_L. C_F



and $R_{\rm ISO}$ serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving the phase margin in the overall feedback loop. Capacitive drive can be increased by increasing the value of C_F . This in turn will slow down the pulse response.

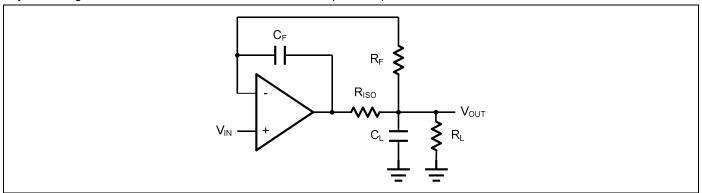


Figure 3. Indirectly Driving a Capacitive Load with DC Accuracy

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Typical Application Circuits

Differential amplifier

The differential amplifier allows the subtraction of two input voltages or cancellation of a signal common the two inputs. It is useful as a computational amplifier in making a differential to single-end conversion or in rejecting a common mode signal. Figure 4. shown the differential amplifier using MCP6001 family

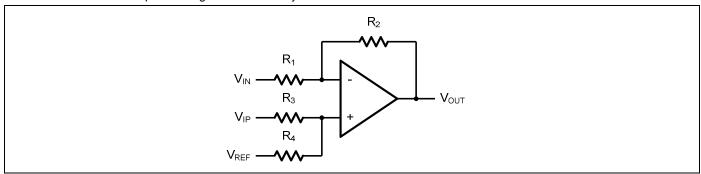


Figure 4. Differential Amplifier

$$V_{\text{OUT}} = \left(\frac{R_1 + R_2}{R_2 + R_4}\right) \frac{R_4}{R_1} V_{\text{IN}} - \frac{R_2}{R_1} V_{\text{IP}} + \left(\frac{R_1 + R_2}{R_2 + R_4}\right) \frac{R_3}{R_1} V_{\text{REF}}$$

If the resistor ratios are equal (i.e. R₁=R₃ and R₂=R₄), then

$$V_{\text{OUT}} = \frac{R_2}{R_1} (V_{\text{IP}} - V_{\text{IN}}) + V_{\text{REF}}$$

Low Pass Active Filter

The low pass active filter is shown in Figure 5. The DC gain is defined by $-R_2/R_1$. The filter has a -20dB/decade roll-off after its corner frequency $f_C=1/(2\pi R_3 C_1)$.

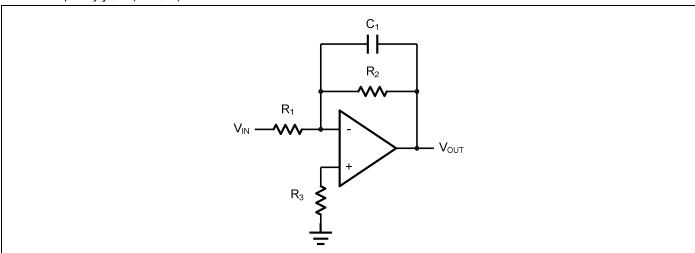


Figure 5. Low Pass Active Filter



Instrumentation Amplifier

The triple MCP6001 family can be used to build a three-op-amp instrumentation amplifier as shown in Figure 6. The amplifier in Figure 6 is a high input impedance differential amplifier with gain of R_2/R_1 . The two differential voltage followers assure the high input impedance of the amplifier.

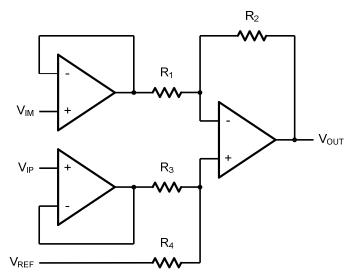
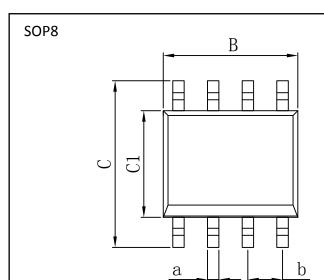
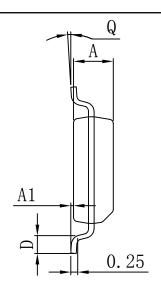


Figure 6. Instrument Amplifier



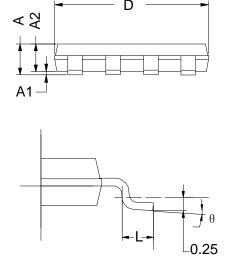
Package Information

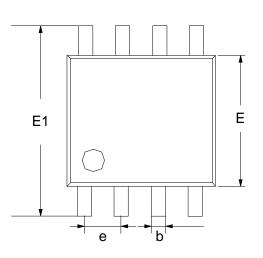




Dimensions In Millimeters						
Symbol:	Min:	Max:	Symbol:	Min:	Max:	
Α	1.225	1.570	D	0.400	0.950	
A1	0.100	0.250	Q	0°	8°	
В	4.800	5.100	а	0.420 TYP		
С	5.800	6.250	b	1.270 TYP		
C1	3.800	4.000		•		

MSOP8

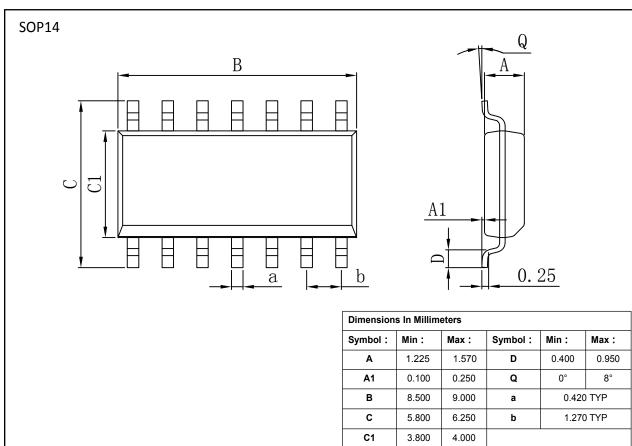




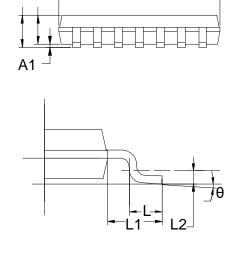
Dimensions In Millimeters						
Symbol:	Min :	Max:	Symbol :	Min :	Max:	
Α	0.800	1.200	E1	4.700	5.100	
A1	0	0.200	L	0.410	0.650	
A2	0.760	0.970	θ	0°	6°	
D	2.900	3.100	b	0.300 TYP		
Е	2.900	3.100	е	0.650 TYP		

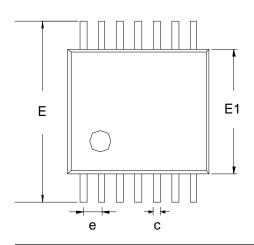


Package Information







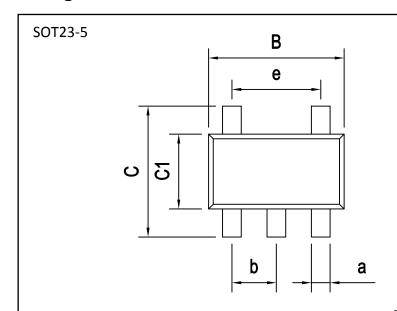


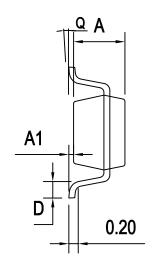
Dimensions In Millimeters						
Symbol :	Min:	Max:	Symbol :	Min:	Max:	
Α	0.950	1.200	E1	4.300	4.500	
A1	0.050	0.150	L	0.450	0.750	
A2	0.800	1.000	θ	0°	8°	
В	0.200	0.280	е	0.650	BSC	
С	0.100	0.190	L1	1.000) REF	
D	4.860	5.060	L2	1.250	BSC	
E	6.200	6.600				

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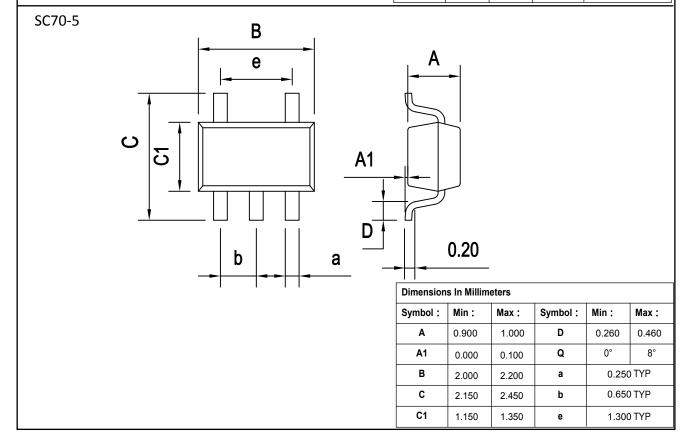


Package Information





Dimensions In Millimeters						
Symbol:	Min :	Max:	Symbol :	Min:	Max:	
Α	1.050	1.150	D	0.300	0.600	
A 1	0.000	0.100	Q	0°	8°	
В	2.820	3.020	а	0.400 TYP		
С	2.650	2.950	b	0.950 TYP		
C1	1.500	1.700	е	1.900 TYP		





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