

OPA277/OPA2277/OPA4277 High Precision Operational Amplifiers

1 Features

Ultralow Offset Voltage: 10 μV
 Ultralow Drift: ±0.1 μV/°C

High Open-Loop Gain: 134 dB

High Common-Mode Rejection: 140 dB
High Power Supply Rejection: 130 dB
Low Bias Current: 1-nA maximum
Wide Supply Range: ±2 V to ±18 V
Low Quiescent Current: 800 μA/amplifier

Single, Dual, and Quad Versions

2 Applications

- Transducer Amplifiers
- Bridge Amplifiers
- Temperature Measurements
- Strain Gage Amplifiers
- · Precision Integrators
- Battery-Powered Instruments
- Test Equipment

3 Description

The OPAx277 series precision operational amplifiers replace the industry standard OPA177. They offer improved noise, wider output voltage swing, and are twice as fast with half the quiescent current. Features include ultralow offset voltage and drift, low bias current, high common-mode rejection, and high power supply rejection. Single, dual, and quad versions have identical specifications, for maximum design flexibility.

OPAx277 series operational amplifiers operate from $\pm 2\text{-V}$ to $\pm 18\text{-V}$ supplies with excellent performance. Unlike most operational amplifiers which are specified at only one supply voltage, the OPAx277 series is specified for real-world applications; a single limit applies over the $\pm 5\text{-V}$ to $\pm 15\text{-V}$ supply range. High performance is maintained as the amplifiers swing to their specified limits. Because the initial offset voltage ($\pm 20~\mu\text{V}$ maximum) is so low, user adjustment is usually not required. However, the single version (OPA277) provides external trim pins for special applications.

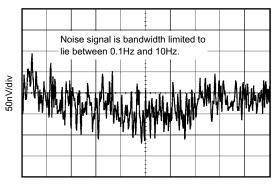
OPA277 operational amplifiers are easy to use and free from phase inversion and the overload problems found in some other operational amplifiers. They are stable in unity gain and provide excellent dynamic behavior over a wide range of load conditions. Dual and quad versions feature completely independent circuitry for lowest crosstalk and freedom from interaction, even when overdriven or overloaded.

Device Information⁽¹⁾

PART NUMBER	PACKAGE	BODY SIZE (NOM)		
	VSON (8)	4.00 mm × 4.00 mm		
OPA277 OPA2277	SOIC (8)	3.91 mm × 4.90 mm		
OTTELLT	PDIP (8)	6.35 mm × 9.81 mm		
ODA 4077	SOIC (14)	3.91 mm × 8.65 mm		
OPA4277	PDIP (14)	6.35 mm × 19.30 mm		

(1) For all available packages, see the orderable addendum at the end of the data sheet.

0.1 Hz to 10 Hz Noise

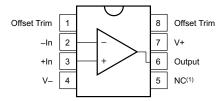


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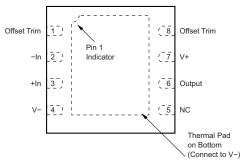


4 Pin Configuration and Functions

OPA277P and D Packages 8-Pin PDIP and SOIC Top View



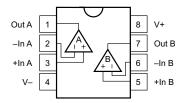
OPA277 DRM Package 8-Pin VSON Top View



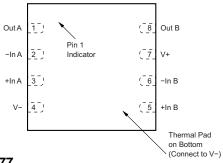
Pin Functions: OPA277

PIN		1/0	DECORPORA			
NO.	NAME	1/0	DESCRIPTION			
1	Offset Trim	I	Input offset voltage trim (leave floating if not used)			
2	–In	I	Inverting input			
3	+In	I	Noninverting input			
4	V-	_	Negative (lowest) power supply			
5	NC	_	No internal connection (can be left floating)			
6	Output	0	Output			
7	V+	_	Positive (highest) power supply			
8	Offset Trim	_	Input offset voltage trim (leave floating if not used)			

OPA2277P and D Packages 8-Pin PDIP and SOIC Top View



OPA2277DRM Package 8-Pin VSON Top View

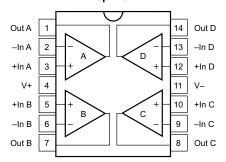


Pin Functions: OPA2277

	PIN						
NAME	PDIP, SOIC NO.	DFN NO.	I/O	DESCRIPTION			
Out A	1	1	0	Output channel A			
−In A	2	2	I	Inverting input channel A			
+In A	3	3	I	Noninverting input channel A			
V-	4	4	_	Negative (lowest) power supply			
+In B	5	5	I	Noninverting input channel B			
−In B	6	6	I	Inverting input channel B			
Out B	7	8	0	Output channel B			
V+	8	7	_	Positive (highest) power supply			



OPA4277P and D Packages 14 Pins PDIP and SOIC Top View



Pin Functions: OPA4277

	PIN							
		1/0	DESCRIPTION					
NO.	NAME		10.00					
1	Out A	0	Output channel A					
2	−In A	I	Inverting input channel A					
3	+In A	I	Noninverting input channel A					
4	V+	_	Positive (highest) power supply					
5	+In B	I	Noninverting input channel B					
6	−In B	I	Inverting input channel B					
7	Out B	0	Output channel B					
8	Out C	0	Output channel C					
9	−In C	I	Inverting input channel C					
10	+In C	I	Noninverting input channel C					
11	V-	_	Negative (lowest) power supply					
12	+In D	I	Noninverting input channel D					
13	–In D	I	Inverting input channel D					
14	Out D	0	Output channel D					



5 Specifications

5.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted) (1)

	MIN	MAX	UNIT
Supply voltage, $Vs = (V+) - (V-)$		36	V
Input voltage	(V-) -0.7	(V+) +0.7	V
Output short-circuit ⁽²⁾	Conti	nuous	
Operating temperature	-55	125	°C
Junction temperature		150	°C
Lead temperature		300	°C
Storage temperature, T _{stq}	-55	125	°C

⁽¹⁾ Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

5.2 ESD Ratings

			VALUE	UNIT
		Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 (1)	±2000	
V _(ESD)	Electrostatic discharge	Charged-device model (CDM), per JEDEC specification JESD22-C101 (2)	±500	V

⁽¹⁾ JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

5.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

	MIN	NOM	MAX	UNIT
Supply voltage, $Vs = (V+) - (V-)$	4 (±2)	30 (±15)	36 (±18)	V
Specified temperature	-40		+85	°C

5.4 Thermal Information for OPA277

			OPA277		
	THERMAL METRIC ⁽¹⁾	P (PDIP)	D (SOIC)	DRM (VSON)	UNIT
			8 PINS		
$R_{\theta JA}$	Junction-to-ambient thermal resistance	49.2	110.1	40.7	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	39.4	52.2	41.3	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	26.4	52.3	16.7	°C/W
ΨЈТ	Junction-to-top characterization parameter	15.4	10.4	0.6	°C/W
Ψ_{JB}	Junction-to-board characterization parameter	26.3	51.5	16.9	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	_	_	3.3	°C/W

For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report, SPRA953.

5.5 Thermal Information for OPA2277

			OPA2277		
	THERMAL METRIC ⁽¹⁾	P (PDIP)	D (SOIC)	DRM (VSON)	UNIT
			8 PINS		
$R_{\theta JA}$	Junction-to-ambient thermal resistance	47.2	107.4	39.3	°C/W
R ₀ JC(top)	Junction-to-case (top) thermal resistance	36.0	45.8	36.9	°C/W

For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report, SPRA953.

⁽²⁾ Short-circuit to ground, one amplifier per package.

⁽²⁾ JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.



Thermal Information for OPA2277(continued)

		OPA2277					
	THERMAL METRIC ⁽¹⁾	P (PDIP)	D (SOIC)	DRM (VSON)	UNIT		
			8 PINS				
$R_{\theta JB}$	Junction-to-board thermal resistance	24.4	47.9	15.4	°C/W		
ΨЈТ	Junction-to-top characterization parameter	13.4	5.7	0.4	°C/W		
Ψ_{JB}	Junction-to-board characterization parameter	24.3	47.3	15.6	°C/W		
R ₀ JC(bot)	Junction-to-case (bottom) thermal resistance	_	_	2.2	°C/W		

5.6 Thermal Information for OPA4277

		OPA	OPA4277			
	THERMAL METRIC ⁽¹⁾	D (SOIC)	P (PDIP)	UNIT		
		14	PINS			
$R_{\theta JA}$	Junction-to-ambient thermal resistance	67.0	66.3	°C/W		
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	24.1	20.5	°C/W		
$R_{\theta JB}$	Junction-to-board thermal resistance	22.5	26.8	°C/W		
Ψ_{JT}	Junction-to-top characterization parameter	2.2	2.1	°C/W		
ψ_{JB}	Junction-to-board characterization parameter	22.1	26.2	°C/W		
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	_	_	°C/W		

⁽¹⁾ For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report, SPRA953.

5.7 Electrical Characteristics for OPAx277P, OPAx277U, and OPAx277xA

At $T_A = 25$ °C, and $R_L = 2 \text{ k}\Omega$, unless otherwise noted

PARAMETER		TEST CONDITIONS		A277PU A2277PU		0	PA277xA PA2277xA PA4277xA		UNIT	
			MIN	TYP ⁽¹⁾	MAX	MIN	TYP (1)	MAX		
OFFSET \	VOLTAGE									
V _{0S}	Input Offset Vol	tage			±10	±20		±20	±50	μV
		OPA277P,U (high-grade, single)				±30				
	Input Offset Voltage Over	OPA2277P,U (high-grade, dual)	T _A = -40°C to 85°C			±50				μV
	Temperature	All PA, UA, Versions							±100	
		AIDRM Versions								
		OPA277P,U (high-grade, single)			±0.1	±0.15				μV/°C
dV _{0S} /dT	Input Offset Voltage Drift	OPA2277P,U (high-grade, dual)	$T_A = -40$ °C to 85°C		±0.1	±0.25				
		All PA, UA, AIDRM Versions						±0.15	±1	
		vs Time			0.2			See (2)		μV/mo
	Input Offset Voltage: (all	vs Power Supply	V _S = ±2 V to ±18 V		±0.3	±0.5		See (2)	±1	
models)	(PSRR)	$T_A = -40$ °C to 85°C			±0.5			±1	μV/V	
	Channel Separa	ation (dual, quad)	DC		0.1			See (2)		μV/V

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⁽¹⁾ $V_S = \pm 15 \text{ V}$

⁽²⁾ Specifications are the same as OPA277P,U.



Electrical Characteristics for OPAx277P,OPAx277U, and OPAx277xA (continued)

At $T_A = 25^{\circ}C$, and $R_L = 2~k\Omega$, unless otherwise noted

	PARAMETER		TEST CONDITIONS	OPA277PU OPA2277PU		OPA277xA OPA2277xA OPA4277xA			UNIT	
				MIN	TYP ⁽¹⁾	MAX	MIN	TYP (1)	MAX	
INPUT BIA	AS CURRENT									
I _B Input Bias Current		int	$T_A = -40^{\circ}C$ to		±0.5	±1		See (2)	±2.8	nA
I _B	input bias curre	er it	85°C			±2			±4	ПА
Ios	Input Offset Current		$T_A = -40^{\circ}C$ to		±0.5	±1		See (2)	±2.8	nA
105			85°C			±2			±4	
NOISE										
	Input Voltage No	oise, f = 0.1 to 10 Hz			0.22			See (2)		μV_{PP}
		f = 10 Hz			12			See (2)		
e _n	Input Voltage	f = 100 Hz			8			See (2)		nV/√Hz
-11	Noise Density	f = 1 kHz			8			See (2)		,
		f = 10 kHz			8			See (2)		
i _n		ensity, f = 1 kHz			0.2			See (2)		pA/√Hz
	LTAGE RANGE		1			Т				
V _{CM}	Common-Mode	Voltage Range		(V-)+2		(V+)-2	See (2)		See (2)	V
CMRR	Common-Mode	Rejection	V _{CM} = (V-) +2 V to (V+) -2 V	130	140		115	See (2)		dB
OWNER	Mix Common Mode Rejection		T _A = -40°C to 85°C	128			115			<u> </u>
INPUT IM	PEDANCE									
	Differential				100 3			See (2)		MΩ pF
	Common-Mode		$V_{CM} = (V-) +2 V$ to $(V+) -2 V$		250 3			See (2)		$G\Omega \parallel pF$
OPEN-LO	OP GAIN									
			$V_{O} = (V-)+0.5 V$ to $(V+)-1.2 V$, $R_{L} = 10 k\Omega$		140			See (2)		
A _{OL}	Open-Loop Volta	age Gain	$V_{O} = (V-)+1.5 V$ to (V+)-1.5 V, $R_{L} = 2 k\Omega$	126	134		See (2)	See (2)		dB
			$V_{O} = (V-)+1.5 V$ to $(V+)-1.5 V$, $R_{L} = 2 k\Omega$	126			See (2)			dB
			$T_A = -40$ °C to 85°C							
FREQUEN	NCY RESPONSE									
GBW	Gain-Bandwidth	Product			1			See (2)		MHz
SR	Slew Rate				0.8			See (2)		V/µs
		0.1%	V _S = ±15 V,		14			See (2)		
	Settling Time	0.01%	G = 1, 10-V Step		16			See (2)		μs
	Overload Recov	ery Time	$V_{IN} \times G = V_{S}$		3			See (2)		μs
THD+N	Total Harmonic	Distortion + Noise	1 kHz, G = 1, V _O = 3.5 Vrms		0.002%			See (2)		



Electrical Characteristics for OPAx277P, OPAx277U, and OPAx277xA(continued)

At T_A = 25°C, and R_L = 2 k Ω , unless otherwise noted

PARAMETER		TEST CONDITIONS	ОРА277РЏ ОРА2277РЏ			OPA277xA OPA2277xA OPA4277xA			UNIT
			MIN	TYP ⁽¹⁾	MAX	MIN	TYP (1)	MAX	
OUTPUT	T								
		$R_L = 10 \text{ k}\Omega$	(V-)+0.5		(V+)-1.2	See (2)		See (2)	V
	Valta va Outavit	$T_A = -40$ °C to +85°C	(V-)+0.5		(V+)-1.2	See (2)		See (2)	
Vo	Voltage Output	$R_L = 2 k\Omega$	(V-)+1.5		(V+)-1.5	See (2)		See (2)	
		$T_A = -40^{\circ}\text{C} \text{ to} +85^{\circ}\text{C}$	(V-)+1.5		(V+)-1.5	See (2)		See (2)	
I _{SC}	Short-Circuit Current			±35			See (2)		mA
C_{LOAD}	Capacitive Load Drive			See (3)					
Zo	Open-loop output impedance	f = 1 MHz		40			See (2)		Ω
POWER	SUPPLY								
Vs	Specified Voltage Range		±5		±15	See (2)		See (2)	V
	Operating Voltage Range		±2		±18	See (2)		See (2)	V
		I _O = 0		±790	±825		See (2)	See (2)	
IQ	Quiescent Current (per amplifier)	$T_A = -40$ °C to 85°C			±900			See (2)	μA
TEMPER	RATURE RANGE	•			,				
	Specified Range		-40		85	See (2)		See (2)	°C
	Operating Range		-55		125	See (2)		See (2)	°C

⁽³⁾ See Typical Characteristics

6.8 Electrical Characteristics for OPAx277AIDRM

At T_A = 25°C, and R_L = 2 k Ω , unless otherwise noted

	PARAMETI	TEST CONDITIONS	OPA277AIDRM OPA2277AIDRM			UNIT	
			MIN	TYP ⁽¹⁾ MAX			
OFFSET VO	LTAGE						
V _{0S}	Input Offset Voltage				±35	±100	μV
		OPA277P,U (high-grade, single)					
	Input Offset Voltage Over Temperature OPA2277P,U (high-grade, dual) All PA, UA, Versions AIDRM Versions		T _A = -40°C to 85°C				μV
		AIDRM Versions				±165	
	Input Offset Voltage Drift Input Offset Voltage: (all models)	OPA277P,U (high-grade, single)	T _A = -40°C to 85°C				
dV _{os} /dT		OPA2277P,U (high-grade, dual)					μV/°C
		All PA, UA, AIDRM Versions			±0.15	±1	
		vs Time			See (2)		μV/mo
		va Dawar Cumhy (DCDD)	$V_S = \pm 2 \text{ V to } \pm 18 \text{ V}$		See (2)	±1	\/\/
	(353510)	vs Power Supply (PSRR)	$T_A = -40$ °C to 85°C			±1	μV/V
	Channel Separation (d	lual, quad)	DC		See (2)		μV/V

⁽¹⁾ $V_S = \pm 15 \text{ V}$ (2) Specifications are the same as OPA277P, U.



Electrical Characteristics for OPAx277AIDRM (continued)

At $T_A = 25^{\circ}C$, and $R_L = 2~k\Omega$, unless otherwise noted

	PARAMETE	ER .	TEST CONDITIONS	OPA277AIDRM OPA2277AIDRM			UNIT	
				MIN	TYP ⁽¹⁾	MAX		
INPUT BIAS C	URRENT							
I _B Input Bias Current			T _A = -40°C to 85°C	±2.8			nA	
						±4		
Ios	Input Offset Current		$T_A = -40$ °C to 85°C	±2.8		±2.8 ±4	nA	
NOISE						14		
NOIGE	Input Voltage Noise, f	= 0.1 to 10 Hz			See (2)		μV _{PP}	
	par rainaga riaisa, r	f = 10 Hz			See (2)		F - FF	
	Input Voltage Noise	f = 100 Hz			See (2)		,	
e _n	Density	f = 1 kHz			See (2)		nV/√Hz	
		f = 10 kHz			See (2)			
i _n	Current Noise Density,	f = 1 kHz			See (2)		pA/√Hz	
INPUT VOLTA	GE RANGE					I		
V _{CM}	Common-Mode Voltag	e Range		See (2)		See (2)	V	
CMRR	Common-Mode Reject	ion	V _{CM} = (V-) +2 V to (V+) -2 V	115	See (2)		dB	
			$T_A = -40$ °C to 85°C	115				
INPUT IMPEDA	ANCE							
	Differential				See (2)		MΩ pF	
	Common-Mode		$V_{CM} = (V-) +2 V \text{ to}$ (V+) -2 V		See (2)		$G\Omega \parallel pF$	
OPEN-LOOP G	BAIN							
	Open-Loop Voltage Gain		$V_O = (V-)+0.5 \text{ V to}$ (V+)-1.2 V, $R_L = 10 \text{ k}\Omega$		See (2)			
A _{OL}			$V_O = (V-)+1.5 \text{ V to}$ (V+)-1.5 V, $R_L = 2 \text{ k}\Omega$	See (2)	See (2)		dB	
			$V_{O} = (V-)+1.5 \text{ V to}$ (V+)-1.5 V, $R_{L} = 2 \text{ k}\Omega$	See ⁽²⁾			dB	
			$T_A = -40$ °C to 85°C					
FREQUENCY F	RESPONSE							
GBW	Gain-Bandwidth Produ	ct			See (2)		MHz	
SR	Slew Rate				See (2)		V/µs	
	Sattling Time	0.1%	$V_S = \pm 15 \text{ V},$		See (2)		110	
	Settling Time	0.01%	G = 1, 10-V Step		See (2)		μs	
	Overload Recovery Tir	ne	$V_{IN} \times G = V_{S}$		See (2)		μs	
THD+N	Total Harmonic Distort	ion + Noise	1 kHz, G = 1, V _O = 3.5 Vrms		See (2)			
OUTPUT								
			$R_L = 10 \text{ k}\Omega$	See (2)		See (2)	.	
Vo	Voltage Output		$T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$	See (2)		See (2)	V	
v O			$R_L = 2 k\Omega$	See (2)		See (2)	V	
			$T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$	See (2)		See (2)		
I _{SC}	Short-Circuit Current				See (2)		mA	
C _{LOAD}	Capacitive Load Drive							
Z _O	Open-loop output impe	edance	f = 1 MHz		See (2)		Ω	



Electrical Characteristics for OPAx277AIDRM (continued)

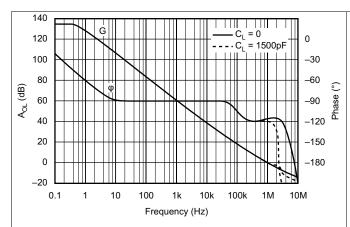
At $T_A = 25^{\circ}C$, and $R_L = 2~k\Omega$, unless otherwise noted

PARAMETER		TEST CONDITIONS	OPA277AIDRM OPA2277AIDRM			UNIT
			MIN	TYP ⁽¹⁾	MAX	
POWER	SUPPLY					
Vs	Specified Voltage Range		See (2)		See (2)	V
	Operating Voltage Range		See (2)		See (2)	V
	Outcome Current (nor amplifier)	I _O = 0		See (2)	See (2)	
IQ	Quiescent Current (per amplifier)	$T_A = -40$ °C to 85°C			See (2)	μΑ
TEMPER	RATURE RANGE					
	Specified Range		See (2)		See (2)	°C
	Operating Range		See (2)		See (2)	°C



6.9 Typical Characteristics

At $T_A = 25$ °C, $V_S = \pm 15$ V, and $R_L = 2$ k Ω , unless otherwise noted.



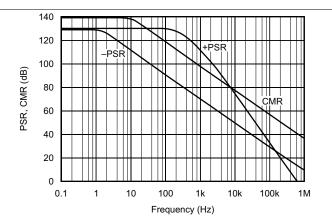
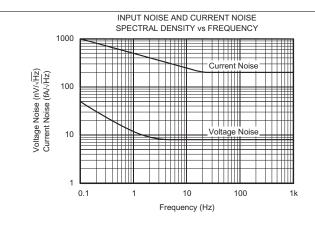


Figure 1. Open-Loop Gain and Phase vs Frequency





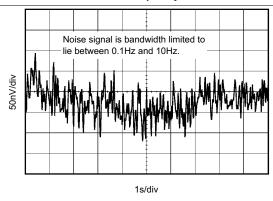
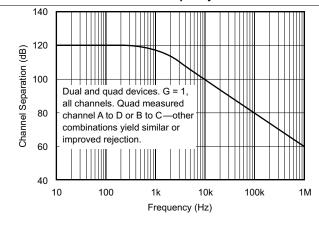


Figure 3. Input Noise and Current Noise Spectral Density vs Frequency

Figure 4. Input Noise Voltage vs Time



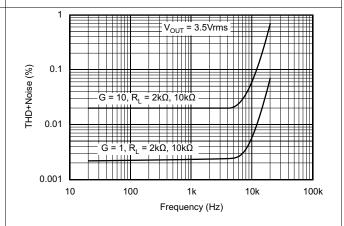
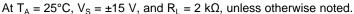


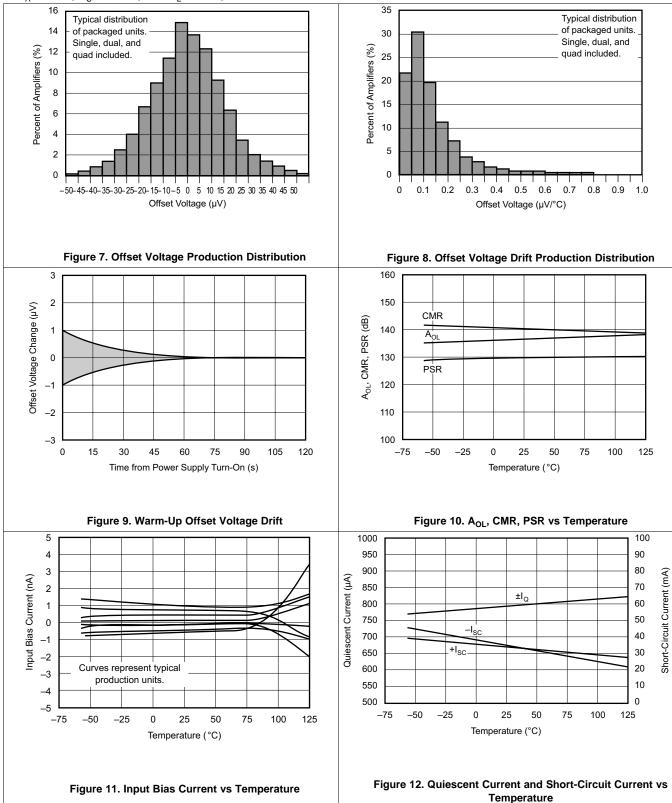
Figure 5. Channel Separation vs Frequency

Figure 6. Total Harmonic Distortion + Noise vs Frequency



Typical Characteristics (continued)



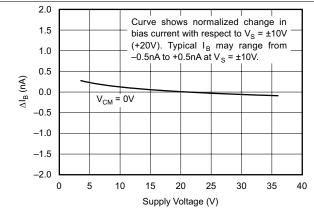


11



Typical Characteristics (continued)

At T_A = 25°C, V_S = ±15 V, and R_L = 2 k Ω , unless otherwise noted.



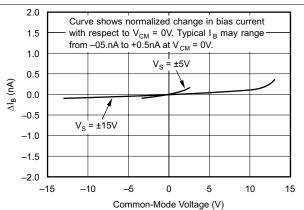
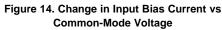
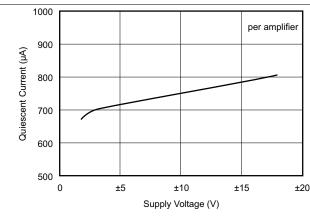


Figure 13. Change in Input Bias Current vs Power Supply Voltage





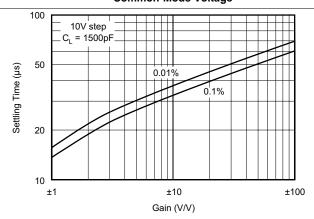
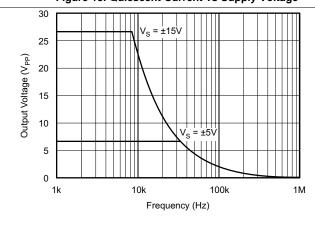


Figure 15. Quiescent Current vs Supply Voltage

Figure 16. Settling Time vs Closed-Loop Gain



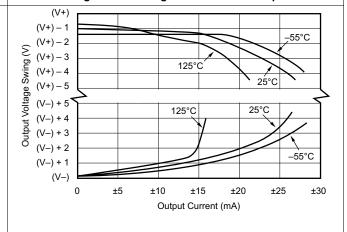


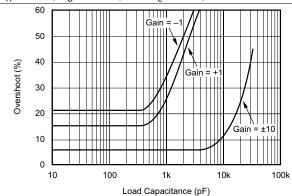
Figure 17. Maximum Output Voltage vs Frequency

Figure 18. Output Voltage Swing vs Output Current



Typical Characteristics (continued)

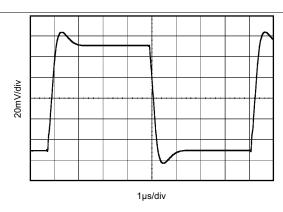
At $T_A = 25$ °C, $V_S = \pm 15$ V, and $R_L = 2$ k Ω , unless otherwise noted.



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Figure 19. Small-Signal Overshoot vs Load Capacitance

Figure 20. Large-Signal Step Response G = 1, $C_L = 1500$ pF, $V_S = \pm 15$ V



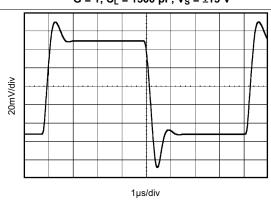


Figure 21. Small-Signal Step Response $G= +1, C_L = 0, V_S = \pm 15 V$

Figure 22. Small-Signal Step Response G= 1, C_L = 1500 pF, V_S = ±15 V

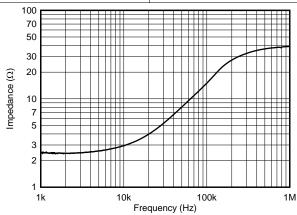


Figure 23. Open-Loop Output Impedance $V_S = \pm 15 \text{ V}$

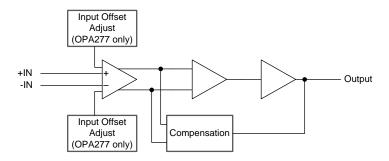


7 Detailed Description

7.1 Overview

The OPAx277 series precision operational amplifiers replace the industry standard OPA177. They offer improved noise, wider output voltage swing, and are twice as fast with half the quiescent current. Features include ultralow offset voltage and drift, low bias current, high common-mode rejection, and high power supply rejection. Single, dual, and quad versions have identical specifications, for maximum design flexibility.

7.2 Functional Block Diagram



7.3 Feature Description

The OPAx277series is unity-gain stable and free from unexpected output phase reversal, making it easy to use in a wide range of applications. Applications with noisy or high-impedance power supplies may require decoupling capacitors close to the device pins. In most cases 0.1-µF capacitors are adequate.

The OPAx277 series has low offset voltage and drift. To achieve highest performance, the circuit layout and mechanical conditions should be optimized. Offset voltage and drift can be degraded by small thermoelectric potentials at the operational amplifier inputs. Connections of dissimilar metals generate thermal potential, which can degrade the ultimate performance of the OPAx277 series. These thermal potentials can be made to cancel by assuring that they are equal in both input terminals.

- Keep the thermal mass of the connections to the two input terminals similar
- Locate heat sources as far as possible from the critical input circuitry
- Shield operational amplifier and input circuitry from air currents, such as cooling fans

7.3.1 Operating Voltage

OPAx277series operational amplifiers operate from $\pm 2\text{-V}$ to $\pm 18\text{-V}$ supplies with excellent performance. Unlike most operational amplifiers, which are specified at only one supply voltage, the OPA277series is specified for real-world applications; a single limit applies over the $\pm 5\text{-V}$ to $\pm 15\text{-V}$ supply range. This allows a customer operating at $V_S = \pm 10\text{ V}$ to have the same assured performance as a customer using $\pm 15\text{-V}$ supplies. In addition, key parameters are assured over the specified temperature range, -40°C to 85°C . Most behavior remains unchanged through the full operating voltage range ($\pm 2\text{ V}$ to $\pm 18\text{ V}$). Parameters which vary significantly with operating voltage or temperature are shown in *Typical Characteristics*.

7.3.2 Offset Voltage Adjustment

The OPAx277series is laser-trimmed for low offset voltage and drift, so most circuits do not require external adjustment. However, offset voltage trim connections are provided on pins 1 and 8. Offset voltage can be adjusted by connecting a potentiometer, as shown in Figure 24. Only use this adjustment to null the offset of the operational amplifier. This adjustment should not be used to compensate for offsets created elsewhere in a system, because this can introduce additional temperature drift.



Feature Description (continued)

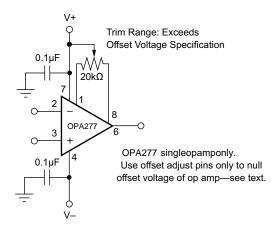


Figure 24. OPA277 Offset Voltage Trim Circuit

7.3.3 Input Protection

The inputs of the OPAx277 series are protected with $1-k\Omega$ series input resistors and diode clamps. The inputs can withstand ± 30 -V differential inputs without damage. The protection diodes conduct current when the inputs are over-driven. This may disturb the slewing behavior of unity-gain follower applications, but will not damage the operational amplifier.

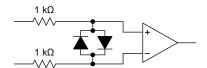


Figure 25. OPAx277 Input Protection

7.3.4 Input Bias Current Cancellation

The input stage base current of the OPAx277series is internally compensated with an equal and opposite cancellation circuit. The resulting input bias current is the difference between the input stage base current and the cancellation current. This residual input bias current can be positive or negative.

When the bias current is canceled in this manner, the input bias current and input offset current are approximately the same magnitude. As a result, it is not necessary to use a bias current cancellation resistor, as is often done with other operational amplifiers (see Figure 26). A resistor added to cancel input bias current errors may actually increase offset voltage and noise.

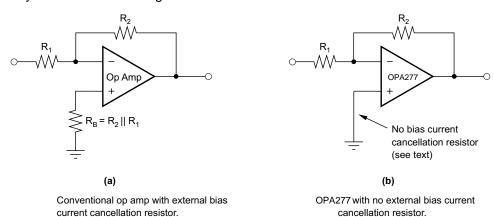


Figure 26. Input Bias Current Cancellation



Feature Description (continued)

7.3.5 EMI Rejection Ratio (EMIRR)

The electromagnetic interference (EMI) rejection ratio, or EMIRR, describes the EMI immunity of operational amplifiers. An adverse effect that is common to many operational amplifiers is a change in the offset voltage as a result of RF signal rectification. An operational amplifier that is more efficient at rejecting this change in offset as a result of EMI has a higher EMIRR and is quantified by a decibel value. Measuring EMIRR can be performed in many ways, but this report provides the EMIRR IN+, which specifically describes the EMIRR performance when the RF signal is applied to the noninverting input pin of the operational amplifier. In general, only the noninverting input is tested for EMIRR for the following three reasons:

- 1. Operational amplifier input pins are known to be the most sensitive to EMI, and typically rectify RF signals better than the supply or output pins.
- 2. The noninverting and inverting operational amplifier inputs have symmetrical physical layouts and exhibit nearly matching EMIRR performance.
- 3. EMIRR is easier to measure on noninverting pins than on other pins because the noninverting input terminal can be isolated on a printed circuit board (PCB). This isolation allows the RF signal to be applied directly to the noninverting input terminal with no complex interactions from other components or connecting PCB traces.

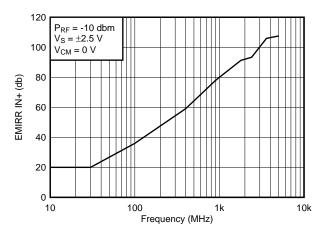


Figure 27. OPA277 EMIRR IN+ vs Frequency

If available, any dual and quad operational amplifier device versions have nearly similar EMIRR IN+ performance. The OPA277unity-gain bandwidth is 1 MHz. EMIRR performance below this frequency denotes interfering signals that fall within the operational amplifier bandwidth.

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Feature Description (continued)

Table 1 shows the EMIRR IN+ values for the OPA277at particular frequencies commonly encountered in real-world applications. Applications listed in Table 1 may be centered on or operated near the particular frequency shown. This information may be of special interest to designers working with these types of applications, or working in other fields likely to encounter RF interference from broad sources, such as the industrial, scientific, and medical (ISM) radio band.

FREQUENCY	APPLICATION/ALLOCATION	EMIRR IN+
400 MHz	Mobile radio, mobile satellite/space operation, weather, radar, UHF	59.1 dB
900 MHz	900 MHz GSM, radio com/nav./GPS (to 1.6 GHz), ISM, aeronautical mobile, UHF	
1.8 GHz	GSM, mobile personal comm. broadband, satellite, L-band	91.3 dB
2.4 GHz	802.11b/g/n, Bluetooth™, mobile personal comm., ISM, amateur radio/satellite, S-band	93.3 dB
3.6 GHz	Radiolocation, aero comm./nav., satellite, mobile, S-band	105.9 dB
5.0 GHz	802.11a/n, aero comm./nav., mobile comm., space/satellite operation, C-band	107.5 dB

Table 1. OPA277 EMIRR IN+ for Frequencies of Interest

7.3.5.1 EMIRR IN+ Test Configuration

Figure 28 shows the circuit configuration for testing the EMIRR IN+. An RF source is connected to the operational amplifier noninverting input terminal using a transmission line. The operational amplifier is configured in a unity gain buffer topology with the output connected to a low-pass filter (LPF) and a digital multimeter (DMM). Note that a large impedance mismatch at the operational amplifier input causes a voltage reflection; however, this effect is characterized and accounted for when determining the EMIRR IN+. The resulting dc offset voltage is sampled and measured by the multimeter. The LPF isolates the multimeter from residual RF signals that may interfere with multimeter accuracy. Refer to SBOA128 for more details.

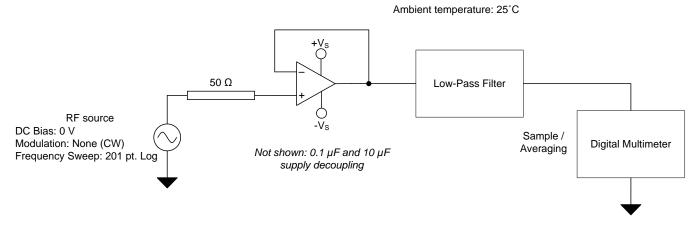


Figure 28. EMIRR IN+ Test Configuration Schematic

7.4 Device Functional Modes

The OPAx277 has a single functional mode and is operational when the power-supply voltage is greater than 4 V (±2 V). The maximum power supply voltage for the OPAx277 is 36 V (±18 V).



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NJM358CG-TE2 HA1630S01LPEL-E LM324AWPT HA1630Q06TELL-E