

RFM42/43 ISM TRANSMITTER Module

Features

- Frequency Range = 240–930 MHz
- Output Power Range
- +8 to +17 dBm (RFM42)
- ◆ -8 to +13 dBm (RFM43)
- Low Power Consumption
- (RFM42)
 - 60 mA @ +17 dBm
 - 27 mA @ +11 dBm
- ◆ (RFM43)
 - 28 mA @ +13 dBm
 - 16 mA @ +1 dBm
- Data Rate = 1 to 128 kbps
- Power Supply = 1.8 to 3.6 V
- Ultra low power shutdown mode
- Wake Up Timer

Applications

- Remote control
- Home security & alarm
- Telemetry
- Personal data logging
- Toy control
- Wireless PC peripherals

- Integrated 32 kHz RC or 32 kHz XTAL
- Integrated voltage regulators
- Configurable packet structure
- TX 64 byte FIFO
- Low battery detector
- Temperature sensor and 8-bit ADC
- -40 to +85 °C temperature range
- Integrated voltage regulators
- Frequency hopping capability
- FSK, GFSK, and OOK modulation
- Power-on-reset (POR)



- Remote keyless entry
- Remote meter reading
- Home automation
- Industrial control
- Sensor networks
- Health monitors

Description

The RFM42/43 is low cost ISM transmitter module and offers advanced radio features including continuous frequency coverage from 240–930 MHz with adjustable power output levels of –8 to +13 dBm on the RFM43 and +8 to +17 dBm on the RFM42. Power adjustments are made in 3 dB steps. The RFM42's Industry leading +17 dBm output power ensures extended range and improved link performance.

Additional system features such as an automatic wake-up timer, low battery detector, 64 byte TX FIFO, and automatic packet handling reduce overall current consumption and allow the use of lower-cost system MCUs. An integrated temperature sensor, general purpose ADC, power-on-reset (POR), and GPIOs further reduce overall system cost and size.

The direct digital transmit modulation and automatic PA power ramping ensure precise transmit modulation and reduced spectral spreading ensuring compliance with FCC and ETSI regulations.

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1. Electrical Specifications

Table 1. DC Characteristics

Parameter	Symbol	Conditions	Min	Тур	Max	Units
Supply Voltage Range	Vdd		1.8	3.0	3.6	V
Power Saving Modes	Shutdown	RC Oscillator, Main Digital Regulator, and Low Power Digital Regulator OFF ²	_	10	TBD	nA
	Standby	Low Power Digital Regulator ON (Register values retained) and Main Digital Regulator, and RC Oscillator OFF ¹	_	400	_	nA
	ISleep	RC Oscillator and Low Power Digital Regulator ON (Register values retained) and Main Digital Regulator OFF ¹	_	800	_	nA
ISensor-L BD ISensor-T S		Main Digital Regulator and Low Battery Detector ON, Crystal Oscillator and all other blocks OFF ²	_	1	_	μA
		Main Digital Regulator and Temperature Sensor ON, Crystal Oscillator and all other blocks OFF ²	_	1		μA
	IReady	Crystal Oscillator and Main Digital Regulator ON, all other blocks OFF. Crystal Oscillator buffer disabled ¹	_	600	_	μA
TUNE Mode Current	Tune	Synthesizer and regulators enabled	_	9.5	_	mA
TX Mode Current for	TX_+17	txpow[2:0] = 011 (+17 dBm), VDD = 3.3 V	_	60	_	mA
RFM42	TX_+11	txpow[2:0] = 000 (+11 dBm), VDD = 3.3 V	—	27	_	mA
TX Mode Current for	TX_+13	txpow[2:0] = 111 (+13 dBm), VDD = 3.3 V	—	28	_	mA
RFM43	TX_+1	txpow[2:0] = 100 (+1 dBm), VDD = 3.3 V	_	16	_	mA

2. Guaranteed by qualification.



Table 2. Synthesizer AC Electrical Characteristics²

Low Band High Band Low Band High Band fcrystal /3	240 480 — —	 156.25 312.5	480 930 —	MHz MHz Hz Hz
Low Band High Band	480 — —		930 —	Hz
High Band				
-		312.5	_	Нz
fcrystal /3	_			112
		10	_	MHz
When using reference frequency instead of crystal. Measured peak-to-peak (VPP)			1.6	V
Measured from leaving Ready mode with XOSC running to any frequency includ-ing VCO Calibration	_	200	_	μs
Integrated over \pm 250 kHz bandwidth (500 Hz lower bound of integration)	—	2	4	kHz rms
△F = 10 kHz	_	-80		dBc/Hz
		-90		dBc/Hz
$\triangle F = 100 \text{ kHz}$		-115		dBc/Hz
$\Delta F = 100 \text{ kHz}$ $\Delta F = 1 \text{ MHz}$		120	_	dBc/Hz
	$\triangle F = 100 \text{ kHz}$	$\triangle F = 100 \text{ kHz}$	$\triangle F = 100 \text{ kHz} \qquad -90$	$\triangle F = 100 \text{ kHz}$ -90 $\triangle F = 1 \text{ MHz}$ -115

1. All specification guaranteed by production test unless otherwise noted.

2. Guaranteed by qualification.



Table 3. Transmitter AC Electrical Characteristics¹

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
rarameter						S
TX Frequency	FSYNTH-LB	Low Band	240	—	480	MHz
Range ¹	Fsynth-hb	High Band	480	—	930	
FSK Modulation Data Rate ²	DRFSK		1	_	128	kbps
OOK Modulation Data Rate ²	DRook		1.2	_	40	kbps
Modulation Deviation ¹	Δf	Production tests maximum limit of 320 kHz	±0.625		±320	kHz
Modulation Deviation Resolution	∆fres		_	0.625	_	kHz
Output Power Range ¹ (RFM43)	Ртх	Power control by txpow[2:0] Register Production test at txpow[2:0] = 11 Tested at 915 MHz	-8	_	+13	dBm
Max Output Power(RFM43)	PTX_max	Tested at 315-915 MHz	+11	+13	_	dBm
Output Power Range ¹ (RFM42)	Ртх	Power control by txpow[2:0] Register Production test at txpow[2:0] = 11 Tested at 915 MHz	+8	_	+17	dBm
Max Output Power(RFM42)	PTX_max	Tested at 315-915 MHz	+15	+17	_	dBm
TX RF Output Steps ²	Prf_out	controlled by txpow[2:0] Register	_	3	_	dB
TX RF Output Level Variation vs. Voltage ²	Prf_v	Measured from VDD=3.6 V to VDD=1.8 V	_	2	_	dB
TX RF Output Level ² Variation vs. Temperature	Prf_temp	40 to +85 ℃	_	2	_	dB
TX RF Output Level Variation vs. Frequency ²	Prf_freq	Measured across any one frequency band	_	1	_	dB
Transmit Modulation Filtering ²	B*T	Gaussian Filtering Bandwith Time Product	_	0.5	_	
Spurious	Ров-тх1	POUT =11dBm, Frequencies <1 GHz			-54	dBm
Emissions ²	Ров-тх2	1–12.75 GHz, excluding harmonics			-54	dBm
Notes:	1		1	I	1	I.

1. All specification guaranteed by production test unless otherwise noted.

2. Guaranteed by qualification.



Table 4. Auxiliary Block Specifications¹

Parameter	Symbol	Conditions	Min	Тур	Max	Units
Temperature Sensor	ΤSA	When calibrated using temp	_	0.5		°C
Accuracy ²	104	sensor offset register		0.0		Ŭ
Temperature Sensor	TSs			5		mV/°C
Sensitivity ²	103			Ŭ		
Low Battery Detector	LBDres		_	50		mV
Resolution ²	LODICES					
Low Battery Detector	LBDct		_	250		μs
Conversion Time ²	LDDCI			200		40
Microcontroller Clock		Configurable to 30 MHz,				
Output Frequency	МС	15 MHz, 10 MHz, 4 MHz,	32.768K	_	30M	Hz
	in o	3 MHz, 2 MHz, 1 MHz, or	02.1 0011		00111	
		32.768 kHz				
General Purpose ADC	ADCENB		_	8		bit
Accuracy ²						
General Purpose ADC	ADCRES		_	4		mV
Resolution ²						
Temp Sensor &						
General						
Purpose ADC	ADC CT		_	305	—	µsec
Conversion						
Time ²						
30 MHz XTAL Start-Up	tзом		_	1	_	ms
time						
30 MHz XTAL Cap	30Mres		_	97		fF
Resolution ²						
32 kHz XTAL Start-Up	t 32K		_	6	_	sec
Time ²	0014					
32 kHz XTAL	32Kres		_	100	_	ppm
Accuracy ²						
32 kHz RC OSC	32KRCres		_	2500	—	ppm
Accuracy ²	1			40		
POR Reset Time	t POR			16		ms
Software Reset Time ² Notes:	t soft			100	—	μs

Notes:

1. All specification guaranteed by production test unless otherwise noted.

2. Guaranteed by qualification.

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Table 5. Digital IO Specifications (SDO, SDI, SCLK, nSEL, and nIRQ)

Parameter	er Symb Conditions		Min	Тур	Мах	Units
	ol					
Rise Time	TRISE	0.1 x VDD to 0.9 x VDD, CL= 5 pF			8	ns
Fall Time	TFALL	0.9 x VDD to 0.1 x VDD, CL= 5 pF	_		8	ns
Input Capacitance	CIN				1	pF
Logic High Level Input Voltage	VIH		VDD – 0.6	—	—	V
Logic Low Level Input Voltage	VIL				0.6	V
Input Current	IIN	0 <vin< td="" vdd<=""><td>-100</td><td>_</td><td>100</td><td>nA</td></vin<>	-100	_	100	nA
Logic High Level Output Voltage	VOH	IOH<1 mA source, VDD=1.8 V	VDD - 0.6		_	V
Logic Low Level Output Voltage	VOL	IOL<1 mA sink, VDD=1.8 V	—	_	0.6	V
Note: All specification guaranteed	by produc	tion test unless otherwise noted.				

Table 6. GPIO Specifications (GPIO_0, GPIO_1, and GPIO_2)

Parameter	Symbol	Conditions	Min	Тур	Max	Units
Rise Time	TRISE	0.1 x VDD to 0.9 x VDD, CL= 10 pF, DRV<1:0>=HH	—		8	ns
Fall Time	TFALL	0.9 x VDD to 0.1 x VDD, CL= 10 pF, DRV<1:0>=HH	_		8	ns
Input Capacitance	CIN		_	_	1	pF
Logic High Level Input Voltage	VIH		VDD – 0.6	—		V
Logic Low Level Input Voltage	VIL		_	_	0.6	V
Input Current	IIN	0 <vin< td="" vdd<=""><td>-100</td><td>_</td><td>100</td><td>nA</td></vin<>	-100	_	100	nA
Input Current If Pullup is Activated	IINP		VIL=0 V	5	_	25
Maximum Output Current	IOmaxLL	DRV<1:0>=LL	0.1	0.5	0.8	mA
	IOmaxLH	DRV<1:0>=LH	0.9	2.3	3.5	mA
	IOmaxHL	DRV<1:0>=HL	1.5	3.1	4.8	mA
	IOmaxHH	DRV<1:0>=HH	1.8	3.6	5.4	mA
Logic High Level Output Voltage	VOH	IOH< IOmax source, VDD=1.8 V	VDD – 0.6		_	V
Logic Low Level Output Voltage	VOL	IOL< IOmax sink, VDD=1.8 V	_	_	0.6	V
Note: All specification guaranteed	by production	n test unless otherwise noted.				



Table 7. Absolute Maximum Ratings

Parameter	Value	Unit				
VDD to GND	-0.3, +3.6	V				
VDD to GND on TX Output Pin	-0.3, +8.0	V				
Voltage on Digital Control Inputs	-0.3, VDD + 0.3	V				
Voltage on Analog Inputs	-0.3, VDD + 0.3	V				
Operating Ambient Temperature Range TA	-40 to +85	°C				
Thermal Impedance θ JA	30	°C/W				
Junction Temperature TJ	+125	°C				
Storage Temperature Range TSTG -55 to +125 °C						
Note: Stresses beyond those listed under "Absolute Maximum Ra	atings" may cause permanent damage to	the device. These				
are stress ratings only and functional operation of the device	are stress ratings only and functional operation of the device at or beyond these ratings in the operational sections of					
the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect						
device reliability. Caution: ESD sensitive device.						
Power Amplifier may be damaged if switched on without proper l	oad or termination connected.					





2. Functional Description

The RFM42/43 is low cost ISM wireless transmitter module with continuous frequency tuning over the complete 240–930 MHz band. The wide operating voltage range of 1.8–3.6 V and low current consumption makes the RFM42/43 and ideal solution for battery powered applications.

A high precision local oscillator (LO) is used for transmit mode. The LO is generated by an integrated VCO and $\triangle \Sigma$ Fractional-N PLL synthesizer. The synthesizer is designed to support configurable data rates, output frequency, frequency deviation, and Gaussian filtering at any frequency between 240–930 MHz. The transmit FSK data is modulated directly into the $\triangle \Sigma$ data stream and can be shaped by a Gaussian low-pass filter to reduce unwanted spectral content. The RFM42's PA output power can be configured between +8 and +17 dBm in 3 dB steps, while the RFM43's PA output

power can be configured between -8 and +13 dBm in 3 dB steps. The PA incorporates automatic ramp-up and ramp-down control to reduce unwanted spectral spreading.

The RFM42/43 is designed to work with a microcontroller to create a very low cost system. Voltage regulators are integrated on-chip which allow for a wide range of operating supply voltage conditions from +1.8 to +3.6 V. A standard 4-pin SPI bus is used to communicate with the microcontroller. Three configurable general purpose I/Os are available for use to tailor towards the needs of the system. A more omplete list of the available GPIO functions is shown in "7. Auxiliary Functions" but just to name a few, microcontroller clock output, POR, and specific interrupts.

2.1. Operating Modes

The RFM42/43 provides several modes of operation which can be used to optimize the power consumption of the device application.

Table 8 summarizes the modes of operation of the RFM42/43. In general, any given mode of operation may be classified as an Active mode or a Power Saving mode. The table indicates which block(s) are enabled (active) in each corresponding mode. With the exception the Shutdown mode, all can be dynamically selected by sending the appropriate commands over the SPI in order to optimize the average current consumption. An "X" in any cell means that, in the given mode of operation, that block can be independently programmed to be either ON or OFF, without noticeably affecting the current consumption. The SPI circuit block includes the SPI interface and the register space. The 32 kHz OSC circuit block includes the 32.768 kHz RC oscillator or 32.768 kHz crystal oscillator, and wake-up timer. AUX (Auxiliary Blocks) includes the temperature sensor, general purpose ADC, and low-battery detector.

Mode	Circuit Blocks								
Name	Digital LDO	SPI	32 kHz OSC	AUX	30 MHz XTAL	PLL	PA	I _{VDD}	
Shutdown	OFF (Register contents lost)	OFF	OFF	OFF	OFF	OFF	OFF	10 nA	
Standby	ON (Register contents retained)	ON	OFF	OFF	OFF	OFF	OFF	400 nA	
Sleep		ON	ON	Х	OFF	OFF	OFF	800 nA	
Sensor		ON	X	ON	OFF	OFF	OFF	1 µA	
Ready		ON	X	Х	ON	OFF	OFF	600 µA	
Tuning		ON	X	Х	ON	ON	OFF	9.5 mA	
Transmit		ON	Х	х	ON	ON	ON	28 mA*	

Table 8. Operating Modes



3. Controller Interface

3.1. Serial Peripheral Interface (SPI)

The RFM42/43 communicates with the host MCU over a 3 wire SPI interface: SCLK, SDI, and nSEL. The host MCU can also read data from internal registers on the SDO output pin. A SPI transaction is a 16-bit sequence which consists of a Read-Write (\overline{R} /W) select bit, followed by a 7-bit address field (ADDR), and an 8-bit data field (DATA), as demonstrated in Figure 1. The 7-bit address field supports reading from or writing to one of the 128, 8-bit control registers. The \overline{R} /W select bit determines whether the SPI transaction is a write or read transaction. If \overline{R} /W = 1, it signifies a WRITE transaction, while \overline{R} /W = 0 signifies a READ transaction. The contents (ADDR or DATA) are latched into the RFM42/43 every eight clock cycles. The timing parameters for the SPI interface are shown in Table 9. The SCLK rate is flexible with a maximum rate of 10 MHz.

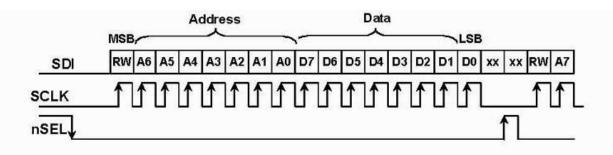


Figure 1. SPI Timing

Table 9. Serial Interface Ti	iming Parameters
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Symbol	Parameter	Min (nsec)	Diagram
t _{CH}	Clock high time	40	
t _{CL}	Clock low time	40	
t _{DS}	Data setup time	20	
t _{DH}	Data hold time	20	
t _{DD}	Output data delay time	20	
t _{EN}	Output enable time	20	
t _{DE}	Output disable time	50	
t _{SS}	Select setup time	20	
t _{SH}	Select hold time	50	
t _{SW}	Select high period	80	

To read back data from the RFM42/43, the \overline{R}/W bit must be set to 0 followed by the 7-bit address of the register from which to read. The 8 bit DATA field following the 7-bit ADDR field is ignored when $\overline{R}/W = 0$. The next eight negative edge transitions of the SCLK signal will clock out the contents of the selected register. The data read from the selected register will be available on the SDO output pin. The READ function is shown in Figure 2. After the READ function is completed the SDO pin will remain at either a logic 1 or logic 0 state depending on the last data bit clocked out (D0). When nSEL goes high the SDO output pin will be pulled high by internal pullup.



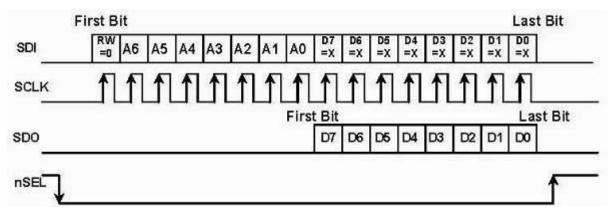


Figure 2. SPI Timing—READ Mode

The SPI interface contains a burst read/write mode which will allows for reading/writing sequential registers without having to re-send the SPI address. When the nSEL bit is held low while continuing to send SCLK pulses, the SPI interface will automatically increment the ADDR and read from/write to the next address. An SPI burst write transaction is demonstrated in Figure 3 and burst read in Figure 2. As long as nSEL is held low, input data will be latched into the RFM42/43 every eight SCLK cycles. A burst read transaction is also demonstrated in Figure 4.

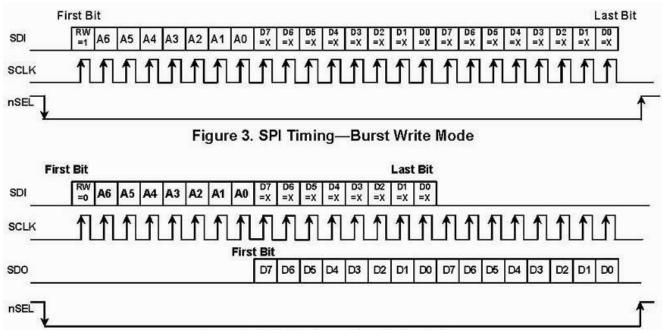


Figure 4. SPI Timing—Burst Read Mode





3.2. Operating Mode Control

There are three primary states in the RFM42/43 radio state machine: SHUTDOWN, IDLE, and TX (see Figure 5). The SHUTDOWN state completely shuts down the radio to minimize current consumption. There are five different configurations/options for the IDLE state which can be selected to optimize the module to the applications needs. "Register 07h. Operating Mode and Function Control 1" controls which operating mode/state is selected. The TX state may be reached automatically from any of the IDLE states by setting the txon bit in "Register 07h. Operating Mode and Function Control 1". Table 10 shows each of the operating modes with the time required to reach TX mode as well as the current consumption of each mode.

The output of the LPLDO is internally connected in parallel to the output of the main digital regulator (and is available externally at the VR_DIG pin); this common digital supply voltage is connected to all digital circuit blocks, including the digital modem, crystal oscillator, and SPI and register space. The LPLDO has extremely low quiescent current consumption but limited current supply capability; it is used only in the IDLE-STANDBY and IDLE-SLEEP modes.

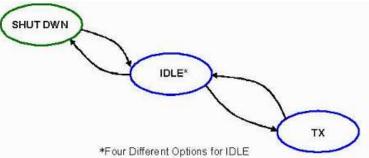


Figure 5. State Machine Diagram

State/Mode	xtal	pll	wt	LBD or TS	Response Time to TX	Current in State /Mode [µA]
Shut Down State	х	х	х	х	16.21 ms	10 nA
Idle States:						
Standby Mode	0	0	0	0	1.21 ms	400 nA
Sleep Mode	0	0	1	0		800 nA
Sensor Mode	0	0	х	1		1 µA
Ready Mode	1	0	х	X	210 µs	600 µA
Tune Mode	1	1	Х	Х	200 µs	9.5 mA
TX State	1	1	х	x	200 µs	RFM42: 80 mA @ +20 dBm, 27 mA @ +11 dBm
						RFM43: 28 mA @ +13 dBm, 16 mA @ +1 dBm

Table 10. Operating Modes



3.2.1. Shutdown State

The shutdown state is the lowest current consumption state of the device with nominally less than 10 nA of current consumption. The shutdown state may be entered by driving the SDN pin high. The SDN pin should be held low in all states except the SHUTDOWN state. In the SHUTDOWN state, the contents of the registers are lost and there is no SPI access.

When the module is connected to the power supply, a POR will be initiated after the falling edge of SDN.

3.2.2. Idle State

There are four different modes in the IDLE state which may be selected by "Register 07h. Operating Mode and Function Control 1". All modes have a tradeoff between current consumption and response time to TX mode. This tradeoff is shown in Table 10. After the POR event, SWRESET, or exiting from the SHUTDOWN state the module will default to the IDLE-READY mode. After a POR event the interrupt registers must be read to properly enter the SLEEP, SENSOR, or STANDBY mode and to control the 32 kHz clock correctly.

3.2.2.1. STANDBY Mode

STANDBY mode has the lowest current consumption possible with only the LPLDO enabled to maintain the register values. In this mode the registers can be accessed in both read and write mode. The standby mode can be entered by writing 0h to "Register 07h. Operating Mode and Function Control 1". If an interrupt has occurred (i.e., the nIRQ pin = 0) the interrupt registers must be read to achieve the minimum current consumption. Additionally, the ADC should not be selected as an input to the GPIO in this mode as it will cause excess current consumption.

3.2.2.2. SLEEP Mode

In SLEEP mode the LPLDO is enabled along with the Wake-Up-Timer, which can be used to accurately wake-up the radio at specified intervals. See "7.6. Wake-Up Timer" for more information on the Wake-Up-Timer. Sleep mode is entered by setting enwt = 1 (40h) in "Register 07h. Operating Mode and Function Control 1". If an interrupt has occurred (i.e., the nIRQ pin = 0) the interrupt registers must be read to achieve the minimum current consumption. Also, the ADC should not be selected as an input to the GPIO in this mode as it will cause excess current consumption.

3.2.2.3. SENSOR Mode

In SENSOR Mode either the Low Battery Detector, Temperature Sensor, or both may be enabled in addition to the LPLDO and Wake-Up-Timer. The Low Battery Detector can be enabled by setting enlbd = 1 and the temperature sensor can be enabled by setting ents = 1 in "Register 07h. Operating Mode and Function Control 1". See "7.4. Temperature Sensor" and "7.5. Low Battery Detector" for more information on these features. If an interrupt has occurred (i.e., the nIRQ pin = 0) the interrupt registers must be read to achieve theminimum current consumption.

3.2.2.4. READY Mode

READY Mode is designed to give a fast transition time to TX mode with reasonable current consumption. In this mode the Crystal oscillator remains enabled reducing the time required to switch to the TX mode by eliminating the crystal start-up time. Ready mode is entered by setting xton = 1 in "Register 07h. Operating Mode and Function Control 1". To achieve the lowest current consumption state the crystal oscillator buffer should be disabled. This is done by setting "Register 62h. Crystal Oscillator/Power-on-Reset Control" to a value of 02h. To exit ready mode, bufovr (bit 1) of this register must be set back to 0.

3.2.2.5. TUNE Mode

In TUNE Mode the PLL remains enabled in addition to the other blocks enabled in the IDLE modes. This will give the fastest response to TX mode as the PLL will remain locked but it results in the highest current consumption. This mode of operation is designed for Frequency Hopping Systems (FHS). Tune mode is entered by setting pllon= 1 in "Register 07h. Operating Mode and Function Control 1". It is not necessary to set xton to 1 for this mode, the internal state machine automatically enables the crystal oscillator.



3.2.3. TX State

The TX state may be entered from any of the IDLE modes when the txon bit is set to 1 in "Register 07h. Operating Mode and Function Control 1". A built-in sequencer takes care of all the actions required to transition between states from enabling the crystal oscillator to ramping up the PA to prevent unwanted spectral splatter. The following sequence of events will occur automatically when going from STANDBY mode to TX mode by setting the txon bit.

- 1. Enable the Main Digital LDO and the Analog LDOs.
- 2. Start up crystal oscillator and wait until ready (controlled by timer).
- 3. Enable PLL.
- 4. Calibrate VCO (this action is skipped when the vcocal bit is "0", default value is "1").
- 5. Wait until PLL settles to required transmit frequency (controlled by timer).
- 6. Activate Power Amplifier and wait until power ramping is completed (controlled by timer).
- 7. Transmit Packet.

The first few steps may be eliminated depending on which IDLE mode the module is configured to prior to setting the txon bit. By default, the VCO and PLL are calibrated every time the PLL is enabled. If the ambient temperature is constant and the same frequency band is being used these functions may be skipped by setting the appropriate bits in "Register 55h. Calibration Control".

3.2.4. Device Status

Add	R/W	Function/Description	D7	D6	D5	D4	D3	D2	D1	D0	POR Def.
02	R	Device Status	ffovfl	ffunfl	Reserved	Reserved			cps[1]	cps[0]	

The operational status of the module can be read from "Register 02h. Device Status".





3.3. Interrupts

The RFM42/43 is capable of generating an interrupt signal when certain events occur. The module notifies the microcontroller that an interrupt event has been detected by setting the nIRQ output pin LOW = 0. This interrupt signal will be generated when any one (or more) of the interrupt events (corresponding to the Interrupt Status bits) shown below occur. The nIRQ pin will remain low until the microcontroller reads the Interrupt Status Register(s) (Registers 03h–04h) containing the active Interrupt Status bit; the nIRQ output signal will then be reset until the next change in status is detected. All of the interrupts must be enabled by the corresponding enable bit in the Interrupt Enable Registers (Registers 05h–06h). All enabled interrupt bits will be cleared when the microcontroller reads the interrupt status register. If the interrupt is not enabled when the event occurs inside of the module it will nottrigger the nIRQ pin, but the status may still be read correctly at anytime in the Interrupt Status registers.

Add	R/W	Function/De	D7	D6	D5	D4	D3	D2	D1	D0	POR
Auu	FK/ VV	scription	יט	00	60	D4	50	DZ	וט	00	Def.
03	R	Interrupt Status 1	ifferr	itxffafull	itxffaem	Reserved	iext	ipksent	Reserved	Reserved	
04	R	Interrupt Status 2	Reserved	Reserved	Reserved	Reserved	iwut	ilbd	ichiprdy	ipor	—
05	R/W	Interrupt Enable1	enfferr	entxffafull	entxffaem	Reserved	enext	enpks ent	Reserved	Reserved	00h
06	R/W	Interrupt Enable 2	Reserved	Reserved	Reserved	Reserved	enwut	enlbd	enchiprdy	enpor	01h

See "Register 03h. Interrupt/Status 1," and "Register 04h. Interrupt/Status 2," for a complete list of interrupts.

3.4. Device Code

The device version code is readable from "Register 01h. Version Code (VC)". This is a read only register.

Add	R/W	Function/Description	D7	D6	D5	D4	D3	D2	D1	D0	POR Def.	Notes
01	R	Device Version	0	0	0	vc[4]	vc[3]	vc[2]	vc[1]	vc[0]	00h	DV



3.5. System Timing

The system timing for TX mode is shown in Figure 6. The timing is shown transitioning from STANDBY mode to TX mode and going automatically through the built-in sequencer of required steps. If a small range of frequencies is being used and the temperature range is fairly constant a calibration may only be needed at the initial power up of the device. The relevant system timing registers are shown below.

۸dd	R/W	Function/De	D7	De	DE	D4	D3	D2	D1	DA	POR
Add	r / w	scription	07	D6	D5	04	03	02	וט	D0	Def.
53	R/W	PLL Tune Time			pllts[4:0]				pllt0[2:0]		45h
54	R/W	Reserved 1	Х	Х	Х	Х	Х	Х	Х	Х	00h
55		Calibration		xtalstart	adccaldo	oprofool	racal	Vcoca	vesel	akinyaa	046
55	R/W	Control		half	ne	enrcfcal	rccal	ldp	vcocal	skipvco	04h

The VCO will automatically calibrate at every frequency change or power up. The VCO CAL may also be forced by setting the vcocal bit. The 32.768 kHz RC oscillator is also automatically calibrated but the calibration may also be forced. The enrcfcal will enable the RC Fine Calibration which will occur every 30 seconds. The rccal bit will force a complete calibration of the RC oscillator which will take approximately 2 ms. The PLL T0 time is to allow for bias settling of the VCO, the default for this should be adequate. The PLL TS time is for the settling time of the PLL, which has a default setting of 200 µs. This setting should be adequate for most applications but may be reduced if small frequency jumps are used. For more information on the PLL register configuration options, see "Register 53h. PLL Tune Time," and "Register 55h. Calibration Control,".

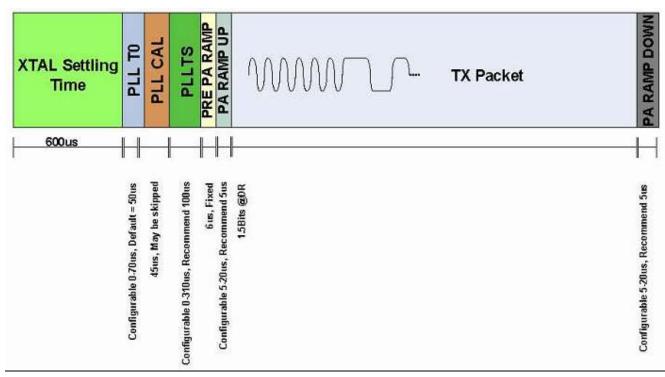


Figure 6. TX Timing



3.6. Frequency Control

3.6.1. Frequency Programming

In order to transmit an RF signal, the desired channel frequency, fcarrier, must be programmed into the RFM42/43.Note that this frequency is the center frequency of the desired channel and not an LO frequency. The carrier frequency is generated by a Fractional-N Synthesizer, using 10 MHz both as the reference frequency and the clock of the (3^{rd} order) $\Delta\Sigma$ modulator. This modulator uses modulo 64000 accumulators. This design was made to obtain the desired frequency resolution of the synthesizer. The overall division ratio of the feedback loop consist of an integer part (N) and a fractional part (F).In a generic sense, the output frequency of the synthesizer is:

fout = $10MHz \times (N + F)$

The fractional part (F) is determined by three different values, Carrier Frequency (fc[15:0]), Frequency Offset (fo[8:0]), and Frequency Modulation (fd[7:0]). Due to the fine resolution and high loop bandwidth of the synthesizer, FSK modulation is applied inside the loop and is done by varying F according to the incoming data; this is discussed further in "3.6.4. Frequency Deviation". Also, a fixed offset can be added to fine-tune the carrier frequency and counteract crystal tolerance errors. For simplicity assume that only the fc[15:0] register will determine the fractional component. The equation for selection of the carrier frequency is shown below:

$$f_{\text{carrier}} = 10MHz \times (hbsel + 1) \times (N + F)$$

$$f_{\text{TX}} = 10MHz * (hbsel+1)* (fb[4:0]+24+\frac{fc[15:0]}{64000})$$

Add	R/W	Function/Description	D7	D6	D5	D4	D3	D2	D1	D0	POR
Auu	FC/ VV	Function/Description	07	00	05	04	03	DZ	וט	00	Def.
73	R/W	Frequency Offset 1	fo[7]	fo[6]	fo[5]	fo[4]	fo[3]	fo[2]	fo[1]	fo[0]	00h
74	R/W	Frequency Offset2							fo[9]	fo[8]	00h
75	R/W	Frequency Band Select		sbsel	hbsel	fb[4]	fb[3]	fb[2]	fb[1]	fb[0]	35h
76	R/W	Nominal Carrier	fo[15]	fo[14]	fo[12]	fo[10]	fc[11	fc[1	fo[0]	fo[0]	BBh
10	K/ W	Frequency 1	fc[15]	fc[14]	fc[13]	fc[12]]	0]	fc[9]	fc[8]	DDII
77	R/W	Nominal Carrier	fo[7]	folGl	folEl	fo[4]	fo[2]	fo[2]	fo[1]	fo[0]	80h
77	FV/VV	Frequency 0	fc[7]	fc[6]	fc[5]	fc[4]	fc[3]	fc[2]	fc[1]	fc[0]	oun

The integer part (N) is determined by fb[4:0]. Additionally, the output frequency can be halved by connecting a $\div 2$ divider to the output. This divider is not inside the loop and is controlled by the hbsel bit in "Register 75h. Frequency Band Select". This effectively partitions the entire 240–930 MHz frequency range into two separate bands: High Band (HB) for hbsel = 1, and Low Band (LB) for hbsel = 0. The valid range of fb[4:0] is from 0 to 23. If a higher value is written into the register, it will default to a value of 23. The integer part has a fixed offset of 24 added to it as shown in the formula above. Table 11 demonstrates the selection of fb[4:0] for the corresponding frequency band.

After selection of the fb (N) the fractional component may be solved with the following equation:

$$fc[15:0] = \left(\frac{f_{TX}}{10MHz * (hbsel + 1)} - fb[4:0]-24\right) * 64000$$

fb and fc are the actual numbers stored in the corresponding registers.



fb[4:0] \/alua	N	Frequen	icy Band		
fb[4:0] Value	IN	hbsel=0	hbsel=1		
0	24	240–249.9 MHz	480–499.9 MHz		
1	25	250–259.9 MHz	500–519.9 MHz		
2	26	260–269.9 MHz	520–539.9 MHz		
3	27	270–279.9 MHz	540–559.9 MHz		
4	28	280–289.9 MHz	560–579.9 MHz		
5	29	290–299.9 MHz	580–599.9 MHz		
6	30	300–309.9 MHz	600–619.9 MHz		
7	31	310–319.9 MHz	620–639.9 MHz		
8	32	320–329.9 MHz	640–659.9 MHz		
9	33	330–339.9 MHz	660–679.9 MHz		
10	34	340–349.9 MHz	680–699.9 MHz		
11	35	350–359.9 MHz	700–719.9 MHz		
12	36	360–369.9 MHz	720–739.9 MHz		
13	37	370–379.9 MHz	740–759.9 MHz		
14	38	380–389.9 MHz	760–779.9 MHz		
15	39	390–399.9 MHz	780–799.9 MHz		
16	40	400–409.9 MHz	800–819.9 MHz		
17	41	410–419.9 MHz	820–839.9 MHz		
18	42	420–429.9 MHz	840–859.9 MHz		
19	43	430–439.9 MHz	860–879.9 MHz		
20	44	440–449.9 MHz	880–899.9 MHz		
21	45	450–459.9 MHz	900–919.9 MHz		
22 46		460–469.9 MHz	920–930.0 MHz		
23	47	470–479.9 MHz	—		

Table 11. Frequency Band Selection

The module will automatically shift the frequency of the Synthesizer down by 937.5 kHz (30 MHz ÷ 32) to achieve the correct Intermediate Frequency (IF).



3.6.2. Easy Frequency Programming for FHSS

While Registers 73h–77h may be used to program the carrier frequency of the RFM42/43, it is often easier to think in terms of "channels" or "channel numbers" rather than an absolute frequency value in Hz. Also, there may be some timing-critical applications (such as for Frequency Hopping Systems) in which it is desirable to change frequency by programming a single register. Once the channel step size is set, the frequency may be changed by a single register corresponding to the channel number. A nominal frequency is first set using Registers 73h–77h, as described above. Registers 79h and 7Ah are then used to set a channel step size and channel number, relative to the nominal setting. The Frequency Hopping Step Size (fhs[7:0]) is set in increments of 10 kHz with a maximum channel step size of 2.56 MHz. The Frequency Hopping Channel Select Register then selects channels based on multiples of the step size.

$F_{carrier} = Fnom + fhs[7:0] X (fhch[7:0] X 10kHz)$

For example: if the nominal frequency is set to 900 MHz using Registers 73h–77h and the channel step size is set to 1 MHz using "Register 7Ah. Frequency Hopping Step Size". For example, if the "Register 79h. Frequency Hopping Channel Select" is set to 5d, the resulting carrier frequency would be 905 MHz. Once the nominal frequency and channel step size are programmed in the registers, it is only necessary to program the fhch[7:0] register in order to change the frequency.

Add	R/W	Function/Descript ion	D7	D6	D5	D4	D3	D2	D1	D0	POR Def.
79	R/W	Frequency Hopping Channel Select	fhch[7]	fhch[6]	fhch[5]	fhch[4]	fhch[3]	fhch [2]	fhch [1]	fhch [0]	00h
7A	R/W	Frequency Hopping Step Size	fhs[7]	fhs[6]	fhs[5]	fhs[4]	fhs[3]	fhs[2]	fhs[1]	fhs[0]	00h

3.6.3. Automatic Frequency Change

If registers 79h or 7Ah are changed in TX mode, the state machine will automatically transition the module back to tune and change the frequency. This feature is useful to reduce the number of SPI commands required in a Frequency Hopping System. This in turn reduces microcontroller activity, reducing current consumption.

3.6.4. Frequency Deviation

The peak frequency deviation is configurable from ± 1 to ± 320 kHz. The Frequency Deviation (Δf) is controlled by the Frequency Deviation Register (fd), address 71 and 72h, and is independent of the carrier frequency setting. When enabled, regardless of the setting of the hbsel bit (high band or low band), the resolution of the frequency deviation will remain in increments of 625 Hz. When using frequency modulation the carrier frequency will deviatefrom the nominal center channel carrier frequency by $\pm \Delta f$:

$$\triangle f = fd [8:0] \times 625Hz$$

$$fd [8:0] = \frac{\triangle f}{625Hz} \triangle \mathbf{f} = \text{peak deviation}$$



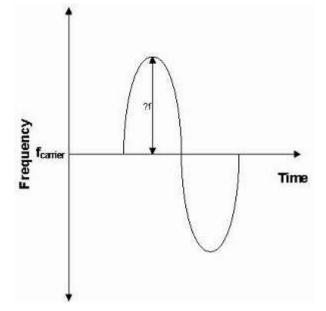


Figure 7. Frequency Deviation

The previous equation should be used to calculate the desired frequency deviation. If desired, frequency modulation may also be disabled in order to obtain an unmodulated carrier signal at the channel center frequency; see "4.1. Modulation Type" for further details.

Add	R/W	Function/Des cription	D7	D6	D5	D4	D3	D2	D1	D0	POR Def.
71	R/W	Modulation Mode Control 2	trclk[1]	trclk[0]	dtmod[1]	dtmod[0]	eninv	fd[8]	modtyp[1]	modtyp[0]	00h
72	R/W	Frequency Deviation	fd [7]	fd [6]	fd [5]	fd [4]	fd [3]	fd [2]	fd [1]	fd [0]	43h



3.6.5. Frequency Offset Adjustment

A frequency offset can be adjusted manually by fo[9:0] in registers 73h and 74h. The frequency offset adjustment is implemented by shifting the Synthesizer Local Oscillator frequency. This register is a signed register so in order to get a negative offset you will need to take the twos complement of the positive offset number. The offset can be calculated by the following:

$$DesiredOffset = 156.25Hz \times (hbse! + 1) \times fo[9:0]$$
$$fo[9:0] = \frac{DesiredOffset}{156.25Hz \times (hbse! + 1)}$$

The adjustment range in high band is: ± 160 kHz, and adjustment range in low band is: ± 80 kHz. For example to compute an offset of +50 kHz in high band mode fo[9:0] should be set to 0A0h. For an offset of -50 kHz in high band mode the fo[9:0] register should be set to 360h.

Add	R/W	Function/Descri ption	D7	D6	D5	D4	D3	D2	D1	D0	POR Def.	Notes
73	R/W	Frequency Offset	fo[7]	fo[6]	fo[5]	fo[4]	fo[3]	fo[2]	fo[1]	fo[1]	00h	73
74	R/W	Frequency Offset							fo[9]	fo[8]	43h	

3.6.6. TX Data Rate Generator

The data rate is configurable between 1–128 kbps. For data rates below 30 kbps the "txdtrtscale" bit in register 70 hshould be set to 1. When higher data rates are used this bit should be set to 0.

The TX date rate is determined by the following formula:

$$DR_TX = \frac{txdr[15:0] \times 1MHz}{2^{16+5*xdtrtscale}}$$

$$txdr[15:0] = \frac{DR_TX \times 2^{16+5*xdtrtscale}}{1MHz}$$

The txdr register may be found in the following registers.

Add	R/W	Function/Des cription	D7	D6	D5	D4	D3	D2	D1	D0	POR Def.
6E	R/W	TX Data Rate 1	txdr[15]	txdr[14]	txdr[13]	txdr[12]	txdr[11]	txdr[10]	txdr[9]	txdr[8]	0Ah
6F	R/W	TX Data Rate 0	txdr[7]	txdr[6]	txdr[5]	txdr[4]	txdr[3]	txdr[2]	txdr[1]	txdr[0]	AAh



4. Modulation Options

4.1. Modulation Type

The RFM42/43 supports three different modulation options: Gaussian Frequency Shift Keying (GFSK), Frequency Shift Keying (FSK), and On-Off Keying (OOK). GFSK is the recommended modulation type as it provides the best performance and cleanest modulation spectrum. Figure 8 demonstrates the difference between FSK and GFSK for a Data Rate of 64 kbps. The time domain plots demonstrate the effects of the Gaussian filtering. The frequency omain plots demonstrate the spectral benefit of GFSK over FSK. The type of modulation is selected with the modtyp[1:0] bits in "Register 71h. Modulation Mode Control 2". Note that it is also possible to obtain an unmodulated carrier signal by setting modtyp[1:0] = 00.

modtyp[1:0]	Modulation Source
00	Unmodulated Carrier
01	OOK
10	FSK
11	GFSK (enable TX Data CLK when direct mode is used)

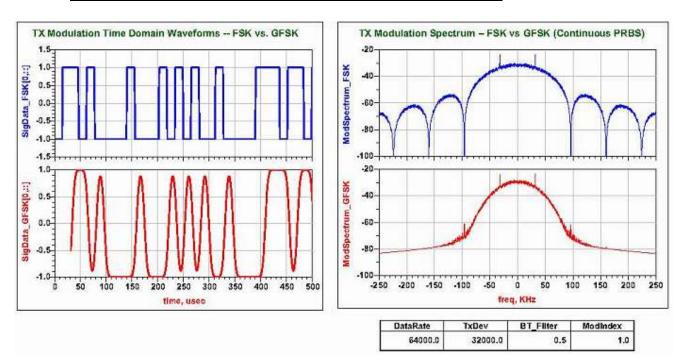


Figure 8. FSK vs GFSK Spectrums



4.2. Modulation Data Source

The RFM42/43 may be configured to obtain its modulation data from one of three different sources: FIFO mode, Direct Mode, and from a PN9 mode. Furthermore, in Direct Mode, the TX modulation data may be obtained from several different input pins. These options are set through the dtmod[1:0] field in "Register 71h. Modulation Mode Control 2".

Add	R/W	Function/Descr iption	D7	D6	D5	D4	D3	D2	D1	D0	POR Def.
71	R/W	Modulation Mode Control 2	trclk[1]	trclk[0]	dtmod [1]	dtmod [0]	eninv	fd[8]	modtyp[1]	modtyp[0]	23h

modtyp[1:0]	Modulation Source
00	Direct Mode using TX_Data via GPIO pin (GPIO needs programming accordingly also)
01	Direct Mode using TX_Data via SDI pin (only when nSEL is high)
10	FIFO Mode
11	PN9 (internally generated)

4.3. FIFO Mode

In FIFO mode, the integrated FIFO is used to transmit the data. The FIFO is accessed via "Register 7Fh. FIFO Access" with burst write capability. The FIFO may be configured specific to the application packet size, etc. (see "6. Data Handling and Packet Handler" for further information).

When in FIFO mode the module will automatically exit the TX State when the *ipksent* interrupt occurs. The module will return to any of the other states based on the settings in "Register 07h. Operating Mode and Function Control 1". For instance, if both the txon and pllon bits are set, the module will transmit all of the contents of the FIFO and the ipksent interrupt will occur. When this event occurs the module will clear the txon bit and return to pllon or Tune Mode. If no other bits are set in register 07h besides txon initially then the module will return to the Idle state.

4.4. Direct Mode

For legacy systems that have packet handling within an MCU or other baseband module, it may not be desirable to use the FIFO. For this scenario, a Direct Mode is provided which bypasses the FIFOs entirely. In Direct Mode, the TX modulation data is applied to an input pin of the module and processed in "real time" (i.e., not stored in a register for transmission at a later time). There are various configurations for choosing which pin is used for the TX Data. Furthermore, an additional input pin is required for the TX Data Clock if GFSK modulation is desired (only the TX Data input pin is required for FSK). Two options for the source of the TX Data are available in the dtmod[1:0] field, and various configurations for the source of the TX Data Clock may be selected through the trclk[1:0] field.

trclk[1:0]	TX Data Clock Configuration
00	No TX Clock (only for FSK)
01	TX Data Clock is available via GPIO (GPIO needs programming accordingly as well)
10	TX Data Clock is available via SDO pin (only when nSEL is high)
11	TX Data Clock is available via the nIRQ pin

The eninv bit in Address 71h will invert the TX Data for testing purposes.



4.5. PN9 Mode

In this mode the TX Data is generated internally using a pseudorandom (PN9 sequence) bit generator. The primary purpose of this mode is for use as a test mode to observe the modulated spectrum without having to load/provide data.

4.6. Synchronous vs. Asynchronous

In Asynchronous mode no clock is used to synchronize the data to the internal modulator. This mode can only be used with FSK. The advantage of this mode that it saves a microcontroller pin because no data clock is required. The disadvantage is that you don't get the clean spectrum and limited BW of GFSK. If Asynchronous FSK is used the TX_DR register should be set to its maximum value.



5. Internal Functional Blocks

This section provides an overview some of the key blocks of the internal radio architecture.

5.1. Synthesizer

An integrated Sigma Delta ($\Sigma\Delta$) Fractional-N PLL synthesizer capable of operating from 240–930 MHz is rovided on-chip. Using a $\Sigma\Delta$ synthesizer has many advantages; it provides large amounts of flexibility in choosing data rate, deviation, channel frequency, and channel spacing. The transmit modulation is applied directly to the loop in the digital domain through the fractional divider which results in very precise accuracy and control over the transmit deviation. The PLL and Δ - Σ modulator scheme is designed to support any desired frequency and channel spacing in the range from 240–930 MHz with a frequency resolution of 156.25 Hz (Low band) or 312.5 Hz (High band). The transmit data rate can be programmed between 1–128 kbps, and the frequency deviation can be programmed between ±1–160 kHz. These parameters may be adjusted via registers as shown in "3.6. Frequency Control".

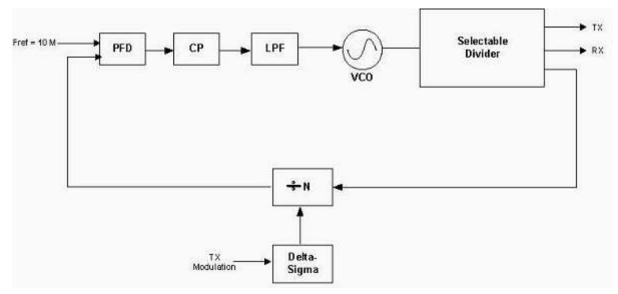


Figure 9. PLL Synthesizer Block Diagram

The reference frequency to the PLL is 10 MHz. The PLL utilizes a differential L-C VCO, with integrated on-chip spiral inductors. The output of the VCO is followed by a configurable divider which will divide down the signal to the desired output frequency band. The modulus of this divider stage is controlled dynamically by the output from the Δ - Σ modulator. The tuning resolution of the Δ - Σ modulator is determined largely by the over-sampling rate and the number of bits carried internally. The tuning resolution is sufficient to tune to the commanded frequency with a maximum accuracy of 312.5 Hz anywhere in the range between 240–930 MHz.

5.1.1. VCO

The output of the VCO is automatically divided down to the correct output frequency depending on the hbsel and fb[4:0] fields in "Register 75h. Frequency Band Select". A 2X VCO is utilized to help avoid problems due to frequency pulling, especially when turning on the integrated Power Amplifier. In receive mode, the LO frequency is automatically shifted downwards (without reprogramming) by the IF frequency of 937.5 kHz, allowing transmit operation on the same frequency. The VCO integrates the resonator inductor, tuning varactor, so no external VCO components are required. The VCO uses capacitance bank to cover the wide frequency range specified. The capacitance bank will automatically be calibrated every time the synthesizer is enabled. In certain fast hopping applications this might not be desirable so the VCO calibration may be skipped by setting the appropriate register.



5.2. Power Amplifier

The RFM42 contains an internal integrated power amplifier (PA) capable of transmitting at output levels between +8 to +17 dBm. The output power is programmable in 3 dB steps through the txpow[2:0] field in "Register 6Dh. TX Power". The RFM43 contains a PA which is capable of transmitting output levels between –8 to +13 dBm.

5.2.1. Output Power Selection

With the RFM42, the output power is configurable in 3 dB steps from +8 to +17 dBm with the txpow[2:0] field in "Register 6Dh. TX Power". The PA output is ramped up and down to prevent unwanted spectral splatter. The higher power setting of the module achieves maximum possible range, but of course comes at the cost of higher TX current consumption. However, depending on the duty cycle of the system, the effect on battery life may be insignificant. Contact HopeRF Support for help in evaluating this tradeoff.

The +13 dBm output power of the RFM43 is targeted at systems that require lower output power. The PA still offers high efficency and a range of output power from –8 to +13 dBm.

Add	R/W	Function/Description	D7	D6	D5	D4	D3	D2	D1	D0	POR Def.
6D	R/W	TX Power						txpow[2]	txpow[1]	txpow[0]	07h

txpow[1:0]	RFM42 Output Power
00	+8 dBm
01	+11 dBm
10	+14 dBm
11	+17 dBm

txpow[2:0]	RFM43 Output Power
000	–8 dBm
001	–5 dBm
010	–2 dBm
011	+1 dBm
100	+4 dBm
101	+7 dBm
110	+10 dBm
111	+13 dBm



5.3. Crystal Oscillator For RF42/RF43 IC

The RF42/43 includes an integrated 30 MHz crystal oscillator with a fast start-up time of less than 600 µs when a suitable parallel resonant crystal is used. The design is differential with the required crystal load capacitance integrated on-chip to minimize the number of external components. By default, all that is required off-chip is the 30 MHz crystal blank.

The crystal load capacitance can be digitally programmed to accommodate crystals with various load capacitance requirements and to slightly adjust the frequency of the crystal oscillator. The tuning of the crystal load capacitance is programmed through the xlc[6:0] field of "Register 09h. 30 MHz Crystal Oscillator Load Capacitance". The total internal capacitance is 12.5 pF and is adjustable in approximately 127 steps (97fF/step). The xtalshift bit is a course shift in frequency but is not binary with xlc[6:0].

The crystal load capacitance can be digitally programmed to accommodate crystals with various load capacitance requirements and to slightly adjust the frequency of the crystal oscillator. This latter function can be used to compensate for crystal production tolerances. Utilizing the on-chip temperature sensor and suitable control software even the temperature dependency of the crystal can be canceled.

The crystal load capacitance is programmed using register 09h. The typical value of the total on-chip (internal) capacitance Cint can be calculated as follows:

Cint = 1.8 pF + 0.085 pF x xlc[6:0] + 3.7 pF x xtalshift

Note that the course shift bit xtalshift is not binary with xlc[6:0]. The total load capacitance Cload seen by the crystal can be calculated by adding the sum of all external parasitic PCB capacitances Cext to Cint. If the maximum value of Cint (16.3 pF) is not sufficient, an external capacitor can be added for exact tuning. See more on this, calculating Cext and crystal selection guidelines in "10. Application Notes".

If AFC is disabled then the synthesizer frequency may be further adjusted by programming the Frequency Offset field fo[9:0]in "Register 73h. Frequency Offset 1" and "Register 74h. Frequency Offset 2", as discussed in "3.6. Frequency Control".

The crystal oscillator frequency is divided down internally and may be output to the microcontroller through one of the GPIO pins for use as the System Clock. In this fashion, only one crystal oscillator is required for the entire system and the BOM cost is reduced. The available clock frequencies (i.e., internal division ratios) and the GPIO configuration are discussed further in "7.2. Microcontroller Clock".

Add	R/W	Function/Descripti on	D7	D6	D5	D4	D3	D2	D1	D0	POR Def.
09	R/W	Crystal Oscillator Load Capacitance	xtalshift	xlc[6]	xlc[5]	xlc[4]	xlc[3]	xlc[2]	xlc[1]	xlc[0]	40h

5.4. Regulators

There are a total of six regulators integrated onto the RFM42/43. With the exception of the IF and Digital all regulators are designed to operate with only internal decoupling. The IF and Digital regulators both require an external 1 μ F decoupling capacitor. All of the regulators are designed to operate with an input supply voltage from +1.8 to +3.6 V, and produce a nominal regulated output voltage of +1.7 V ±5%. The internal circuitry nominally operates from this regulated +1.7 V supply. The output stage of the of PA is not connected internally to a regulator and is connected directly to the battery voltage.

A supply voltage should only be connected to the VDD pins. No voltage should be forced on the IF or DIG regulator outputs.



6. Data Handling and Packet Handler

6.1. TX FIFO

A 64 byte FIFO is integrated into the module for TX, as shown in Figure 10. "Register 7Fh. FIFO Access" is used to access the FIFO. A burst write, as described in "3.1. Serial Peripheral Interface (SPI)", to address 7Fh will write data to the TX FIFO.

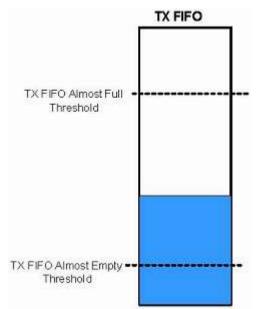


Figure 10. FIFO Threshold

The TX FIFO has two programmable thresholds. An interrupt event occurs when the data in the TX FIFO reaches these thresholds. The first threshold is the FIFO Almost Full threshold, txafthr[5:0]. The value in this register corresponds to the desired threshold value in number of bytes. When the data being filled into the TX FIFO reaches this threshold limit, an interrupt to the microcontroller is generated so the module can enter TX mode to transmit the contents of the TX FIFO. The second threshold for TX is the FIFO Almost Empty Threshold, txaethr[5:0]. When the data being shifted out of the TX FIFO reaches the Almost Empty threshold an interrupt will be generated. The microcontroller will need to switch out of TX mode or fill more data into the TX FIFO. The Transmitter may be configured so that when the TX FIFO is empty the chip will automatically move to the Ready state. In this mode the TX FIFO Almost Empty Threshold may not be useful. This functionality is set by the flidle bit in "Register 08h. Operating Mode and Function Control 2,".





A	dd	R/W	Function/Descri ption	D7	D6	D5	D4	D3	D2	D1	D0	POR Def.
	08	R/W	Operating &Function Control 2	Reser ved	Reser ved	Reserve d	Reserve d	autotx	Reserve d	Reserved	ffclrtx	00h
	7C	R/W	TX FIFO Control 1			txafthr[5]	txafthr[4]	txafthr[3]	txafthr[2]	txafthr[1]	txafthr[0]	37h
	7D	R/W	TX FIFO Control 2			txafthr[5]	txafthr[4]	txafthr[3]	txafthr[2]	txafthr[1]	txafthr[0]	04h

The TX FIFO may be cleared or reset with the ffclrtx bit in "Register 08h. Operating Mode and Function Control 2,". All interrupts may be enabled by setting the Interrupt Enabled bits in "Register 05h. Interrupt Enable 1" and "Register 06h. Interrupt Enable 2,". If the interrupts are not enabled the function will not generate an interrupt on the nIRQ pin but the bits will still be read correctly in the Interrupt Status registers.

6.2. Packet Configuration

When using the FIFO, automatic packet handling may be enabled for the TX mode. "Register 30h. Data Access Control" through "Register 3Eh. Packet Length," control the configuration for Packet Handling. The usual fields for network communication (such as preamble, synchronization word, headers, packet length, and CRC) can be configured to be automatically added to the data payload. The fields needed for packet generation normally change infrequently and can therefore be stored in registers. Automatically adding these fields to the data payload greatly reduces the amount of communication between the microcontroller and the RFM42/43 and therefore also reduces the required computational power of the microcontroller.

The general packet structure is shown in Figure 11. The length of each field is shown below the field. The preamble pattern is always a series of alternating ones and zeroes, starting with a one. All the fields have programmable lengths to accommodate different applications. The most common CRC polynominals are available for selection.

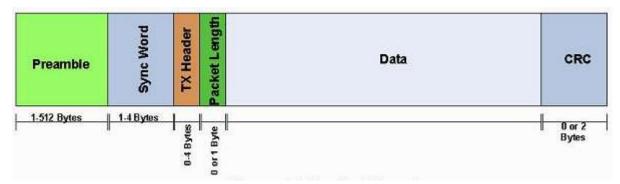


Figure 11. Packet Structure

An overview of the packet handler configuration registers is shown in Table 12. A complete register description can be found in "11.1. Complete Register Table and Descriptions".

6.3. Packet Handler TX Mode

If the TX packet length is set the packet handler will send the number of bytes in the packet length field before returning to ready mode and asserting the packet sent interrupt. To resume sending data from the FIFO the microcontroller needs to command the module to re-enter TX mode Figure 12 provides an example transaction where the packet length is set to three bytes.

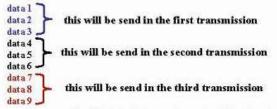


Figure 12. Multiple Packets in TX Packet Handler

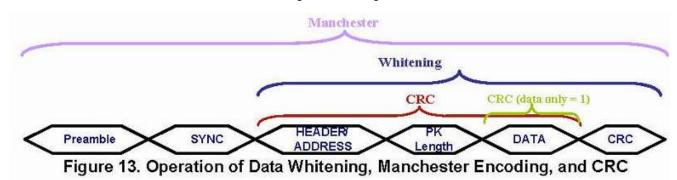
Add	R/W	Function/Description	D7	D6	D5	D4	D3	D2	D1	D0	POR Def.
30	R/W	Data Access Control	enpacix	Isbfrst	crcdonly	*	enpactx	encrc	crc[1]	crc[0]	1Dh
31	R	EzMAC status	0	Reserved	Reserved	Reserved	Reserved	Reserved	pktx	pksent	-
32	RAV	Header Control 1	bcen[3]	enbcast[2]	enbcast[1]	enbcast[0]	Reserved	Reserved	Reserved	Reserved	0Ch
33	R/W	Header Control 2	Reserved	hdlen[2]	hdlen[1]	hdlen[0]	fixpklen	synclen[1]	synclen[0]	prealen[8]	22h
34	RAV	Preamble Length	prealen(7)	prealen(6)	prealen(5)	prealen[4]	prealen[3]	prealen[2]	prealen[1]	prealen(0)	07h
35	R/W	Preamble Detection Control	preath[4]	preath[3]	preath[2]	preath[1]	preath[0]	Reserved	Reserved	Reserved	20h
36	R/W	Sync Word 3	sync[31]	sync[30]	sync[29]	sync[28]	sync[27]	sync[26]	sync[25]	sync[24]	2Dh
37	R/W	Sync Word 2	sync[23]	sync[22]	sync[21]	sync[20]	sync[19]	sync[18]	sync[17]	sync[16]	D4h
38	R/W	Sync Word 1	sync[15]	sync[14]	sync[13]	sync[12]	sync[11]	sync[10]	sync[9]	sync[8]	00h
39	R/W	Sync Word 0	sync[7]	sync[6]	sync[5]	sync[4]	sync[3]	sync[2]	sync[1]	sync[0]	00h
ЗА	RAV	Transmit Header 3	txhd[31]	txhd[30]	txhd[29]	txhd[28]	txhd[27]	txhd[26]	txhd[25]	txhd[24]	00h
3B	R/W	Transmit Header 2	txhd[23]	txhd[22]	txhd[21]	txhd[20]	txhd[19]	txhd[18]	txhd[17]	txhd[16]	00h
30	R/W	Transmit Header 1	txhd[15]	txhd[14]	txhd[13]	txhd[12]	txhd[11]	txhd[10]	txhd[9]	txhd[8]	00h
3D	RAW	Transmit Header 0	txhd[7]	txhd[6]	txhd[5]	txhd[4]	txhd[3]	txhd[2]	txhd[1]	txhd[0]	00h
3E	RAV	Transmit Packet Length	pklen[7]	pklen[6]	pklen[5]	pklen[4]	pklen[3]	pklen[2]	pklen[1]	pklen[0]	00h

Table 12. Packet Handler Registers



6.4. Data Whitening, Manchester Encoding, and CRC

Data whitening can be used to avoid extended sequences of 0s or 1s in the transmitted data stream to achieve a more uniform spectrum. When enabled, the payload data bits are XORed with a pseudorandom sequence output from the built-in PN9 generator. The generator is initialized at the beginning of the payload. The receiver recovers the original data by repeating this operation. Manchester encoding can be used to ensure a dc-free transmission and good synchronization properties. When Manchester encoding is used, the effective datarate is unchanged but the actual datarate (preamble length, etc.) is doubled due to the nature of the encoding. The effective datarate when using Manchester encoding is **limited to 64 kbps**. Data Whitening and Manchester encoding can be selected with "Register 70h. Modulation Mode Control 1". The CRC is configured via "Register 30h. Data Access Control".



6.5. TX Retransmission and Auto TX

The RFM42/43 is capable of automatically retransmitting the last packet in the FIFO if no additional packets were loaded into the TX FIFO. Automatic Retransmission is achieved by entering the TX state with the txon bit set. This feature is useful for Beacon transmission or when retransmission is required due to the absence of a valid acknowledgement. Only packets that fit completely in the TX FIFO are valid for retransmit. When it is necessary to transmit longer packets, the TX FIFO uses the circular read/write capability.

An Automatic Transmission is also available. When autotx = 1 the transceiver will enter automatically TX State when the TX FIFO is almost full. When the TX FIFO is empty the transceiver will automatically return to the IDLE State.

Add	R/W	Function/Description	D7	D6	D5	D4	D3	D2	D1	D0	POR Def.
08	R/W	Operating &Function Control 2		Rese	erved		autotx	Reserved	Reserved	ffclrtx	00h



7. Auxiliary Functions

7.1. Smart Reset

The RFM42/43 contains an enhanced integrated SMART RESET or POR circuit. The POR circuit contains both a classic level threshold reset as well as a slope detector POR. This reset circuit was designed to produce reliable reset signal in any circumstances. Reset will be initiated if any of the following conditions occur:

- Initial power on, when VDD starts from 0V: reset is active till VDD reaches VRR (see table);
- When VDD decreases below VLD for any reason: reset is active till VDD reaches VRR again;
- A software reset via "Register 08h. Operating Mode and Function Control 2,": reset is active for time TSWRST
- On the rising edge of a VDD glitch when the supply voltage exceeds the following time functioned limit:

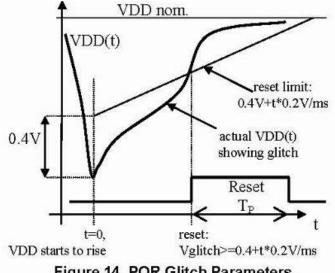


Figure 14. POR Glitch Parameters

Parameter	Symbol	Comment	Min	Тур	Max	Unit
Release Reset Voltage	VRR		0.85	1.3	1.75	v
Power-On VDD Slope	SVDD	tested VDD slope region	0.03		300	V/ms
Low VDD Limit	VLD	VLD <vrr guaranteed<="" is="" td=""><td>0.7</td><td>1</td><td>1.3</td><td>v</td></vrr>	0.7	1	1.3	v
Software Reset Pulse	TSWRST		50		470	us
Threshold Voltage	VTSD			0.4		v
Reference Slope	k			0.2		V/ms
VDD Glitch Reset Pulse	TP	Also occurs after SDN, and initial power on	5	15	40	ms

Table 13. POR Parame	eters
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The reset will initialize all registers to their default values. The reset signal is also available for output and use by the microcontroller by using the default setting for GPIO_0. The inverted reset signal is available by default on GPIO_1.



7.2. Microcontroller Clock

The crystal oscillator frequency is divided down internally and may be output to the microcontroller through GPIO2. This feature is useful to lower BOM cost by using only one crystal in the system. The system clock frequency is selectable from one of 8 options, as shown below. Except for the 32.768 kHz option, all other frequencies are derived by dividing the Crystal Oscillator frequency. The 32.768 kHz clock signal is derived from an internal RC Oscillator or an external 32 kHz Crystal, depending on which is selected. The GPIO2 default is the microcontroller clock with a 1 MHz microcontroller clock output.

Ac	d R	R/W	Function/Descript ion	D7	D6	D5	D4	D3	D2	D1	D0	POR Def.
Α) R	R/W	Microcontroller Output Clock			clkt[1]	clkt[0]	enlfc	mclk[2]	mclk[1]	mclk[0]	00h

mclk[2:0]	Modulation Source
000	30 MHz
001	15 MHz
010	10 MHz
011	4 MHz
100	3 MHz
101	2 MHz
110	1 MHz
111	32.768 KHz

If the microcontroller clock option is being used there may be the need of a System Clock for the microcontroller while the RFM42/43 is in SLEEP mode. Since the Crystal Oscillator is disabled in SLEEP mode in order to save current, the low-power 32.768 kHz clock can be automatically switched to become the microcontroller clock. This feature is called Enable Low Frequency Clock and is enabled by the enlfc bit. When enlfc = 1 and the module is in SLEEP mode then the 32.768 kHz clock will be provided to the microcontroller as the System Clock, regardless of the setting of mclk[2:0]. For example, if mclk[2:0] = 000, 30 MHz will be provided through the GPIO output pin to the microcontroller as the System Clock in all IDLE or TX states. When the module is commanded to SLEEP mode, the System Clock will become 32.768 kHz.

Another available feature for the microcontroller clock is the Clock Tail, clkt[1:0]. If the Enable Low Frequency Clock feature is not enabled (enlfc = 0), then the System Clock to the microcontroller is disabled in SLEEP mode. However, it may be useful to provide a few extra cycles for the microcontroller to complete its operation prior to the shutdown of the System Clock signal. Setting the clkt[1:0] field will provide additional cycles of the System Clock before it shuts off.

clkt[1:0]	Modulation Source				
00	0 cycles				
01	128 cycles				
10	256 cycles				
11	512 cycles				

If an interrupt is triggered, the microcontroller clock will remain enabled regardless of the selected mode. As soon as the interrupt is read the state machine will then move to the selected mode. For instance, if the module is commanded to Sleep mode but an interrupt has occurred the 30 MHz XTAL will not disable until the interrupt has been cleared.



7.3. General Purpose ADC

An 8-bit SAR ADC is integrated onto the module for general purpose use, as well as for digitizing the temperature sensor reading. "Register 0Fh. ADC Configuration," must be configured depending on the use of the GP ADC before use. The architecture of the ADC is demonstrated in Figure 15. First the input of the ADC must be selected by setting the ADCSEL[2:0] depending on the use of the ADC. For instance, if the ADC is going to be used to read out the internal temperature sensor, then ADCSEL[2:0] should be set to 000. Next, the input reference voltage to the ADC must be chosen. By default, the ADC uses the bandgap voltage as a reference so the input range of the ADC is from 0–1.02 V with an LSB resolution of 4 mV (1.02/255). Changing the ADC reference will change the LSB resolution accordingly.

Every time the ADC conversion is desired, the ADCStart bit in "Register 0Fh. ADC Configuration," must be set to 1. This is a self clearing bit that will be cleared at the end of the conversion cycle of the ADC. The conversion time for the ADC is 350 us. After the 350 us or when the ADCstart/busy bit is cleared, then the ADC value may be read out of "Register 11h. ADC Value". Setting the "Register 10h. ADC Sensor Amplifier Offset", ADC Sensor Amplifier Offset is only necessary when the ADC is configured to used as a Bridge Sensor as described in the following section.

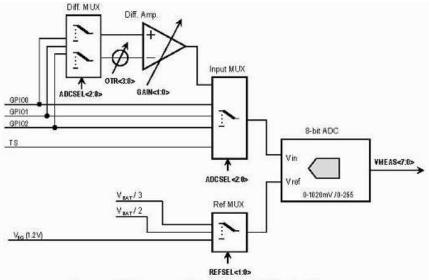


Figure 15. General Purpose ADC Architecture

Add	R/W	Function/Description	D7	D6	D5	D4	D3	D2	D1	DO	POR Def.
0F	R/W	ADC Configuration	adcstart/adcbusy	adcsel[2]	adcsel[1]	adcsel[0]	adcref(1)	adcref[0]	adcgain[1]	adcgain(0)	00h
10	R/W	ADC Sensor Amplifier Offset					adcoffs[3]	adcoffs[2]	adcoffs[1]	adcoffs[0]	00h
11	R	ADC Value	adc[7]	adc[6]	adc[5]	adc[4]	adc[3]	adc[2]	adc[1]	adc[0]	-



7.3.1. ADC Differential Input Mode—Bridge Sensor Example

The differential input mode of ADC8 is designed to directly interface any bridge-type sensor, which is demonstrated in the figure below. As seen in the figure the use of the ADC in this configuration will utilize two GPIO pins. The supply source of the bridge and module should be the same to eliminate the measuring error caused by battery discharging. For proper operation one of the VDD dependent references (VDD/2 or VDD/3) should be selected for the reference voltage of ADC8. VDD/2 reference should be selected for VDD lower than 2.7 V, VDD/3 reference should be selected for VDD higher than 2.7 V. The differential input mode supports programmable gain to match the input range of ADC8 to the characteristic of the sensor and VDD proportional programmable offset adjustment to compensate the offset of the sensor.

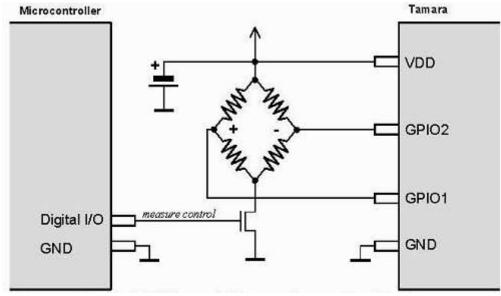


Figure 16. ADC Differential Input Example—Bridge Sensor

The adcgain[1:0] bits in "Register 0Eh. I/O Port Configuration" determine the gain of the differential/single ended amplifier. This is used to fit the input range of the ADC8 to bridge sensors having different sensitivity:

adcgain[1]	adcgain[0]	Differential Gai	n	Input Range (% of VDD)			
		adcref[0] = 0	adcref[0] = 1				
0	0	22/13	33/13	16.7			
0	1	44/13	66/13	8.4			
1	0	66/13	99/13	5.6			
1	1	88/13	132/13	4.2			

Note: The input range is the differential voltage measured between the selected GPIO pins corresponding to the full ADC range (255).

The gain is different for different VDD dependent references so the reference change has no influence on input range and digital measured values.



The differential offset can be coarse compensated by the adcoffs[3:0] bits found in "Register 11h. ADC Value". Fine compensation should be done by the microcontroller software. The main reason for the offset compensation is to shift the negative offset voltage of the bridge sensor to the positive differential voltage range. This is essential as the differential input mode is unipolar. The offset compensation is VDD proportional, so the VDD change has no influence on the measured value.

adcoffs[3]	Input Offset (% of VDD)
0	0 if adcoffs[2:0] = 0
	-(8 - adcoffs[2:0]) x 0.12
1	adcoffs[2:0] x 0.12

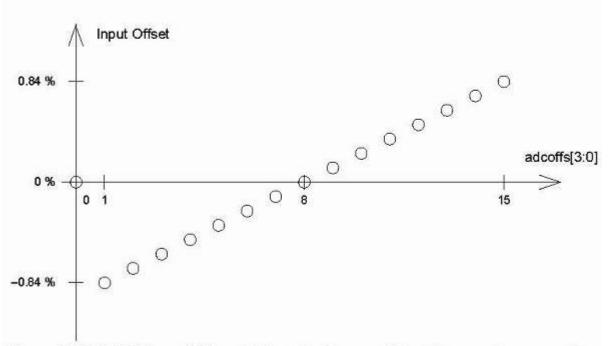


Figure 17. ADC Differential Input Offset for Sensor Offset Coarse Compensation



7.4. Temperature Sensor

An analog temperature sensor is integrated into the module. The temperature sensor will be automatically enabled when the temperature sensor is selected as the input of the ADC or when the analog temp voltage is selected on the analog test bus. The temperature sensor value may be digitized using the general-purpose ADC and read out over the SPI through "Register 10h. ADC Sensor Amplifier Offset". The range of the temperature sensor is selectable to configure to the desired application and performance. The table below demonstrates the settings for the different temperature ranges and performance.

To use the Temp Sensor:

- 1. Set input for ADC to be Temperature Sensor, "Register 0Fh. ADC Configuration"—adcsel[2:0] = 000
- 2. Set Reference for ADC, "Register 0Fh. ADC Configuration"-adcref[1:0] = 00
- 3. Set Temperature Range for ADC, "Register 12h. Temperature Sensor Calibration"-tsrange[1:0]
- 4. Set entsoffs = 1, "Register 12h. Temperature Sensor Calibration"
- 5. Trigger ADC Reading, "Register 0Fh. ADC Configuration"—adcstart = 1

Add	R/W	Function/Descr iption	D7	D6	D5	D4	D3	D2	D1	D0	POR Def.
12	R/W	Temperature Sensor Control	tsrange[1]	tsrange[0]	entsoffs	entstrim	vbgtrim[3]	vbgtrim[2]	vbgtrim[2]	vbgtrim[2]	20h
13	R/W	Temperature Value Offset	tvoffs[7]	tvoffs[7]	tvoffs[7]	tvoffs[7]	tvoffs[7]	tvoffs[7]	tvoffs[7]	tvoffs[7]	00h

6. Read-out Value—Read Address in "Register 11h. ADC Value"

Table 14. Temperature Sensor Range

entoff	tsrange[1]	tsrange[0]	Temp. range	Unit	Slope	ADC8 LSB					
1	0	0	–64 … 64 °C	°C	8 mV/°C	0.5 °C					
1	0	1	-64 192	°C	4 mV/°C	1 °C					
1	1	0	0 128	°C	8 mV/°C	0.5 °C					
1	1	1	-40 216	°F	4 mV/°F	1 °F					
0* 1 0 0 341 °K 3 mV/°K 1.333 °K											
*Note: Ab	*Note: Absolute temperature mode, no temperature shift. This mode is only for test purposes. POR value of										

EN TOFF is 1.

Control to adjust the temperature sensor accuracy is available by adjusting the bandgap voltage. By enabling the envbgcal and using the vbgcal[3:0] bits to trim the bandgap the temperature sensor accuracy may be fine tuned in the final application. The slope of the temperature sensor is very linear and monotonic but the exact accuracy or offset in temperature is difficult to control better than ± 10 °C. With the vbgtrim or bandgap trim though the initial temperature offset can be easily adjusted and be better than ± 3 °C.

The different ranges for the temperature sensor and ADC8 are demonstrated in Figure 18. The value of the ADC8 may be translated to a temperature reading by ADC8Value x ADC8 LSB + Lowest Temperature in Temp Range. For instance for a tsrange = 00, Temp = ADC8Value x 0.5 - 64.



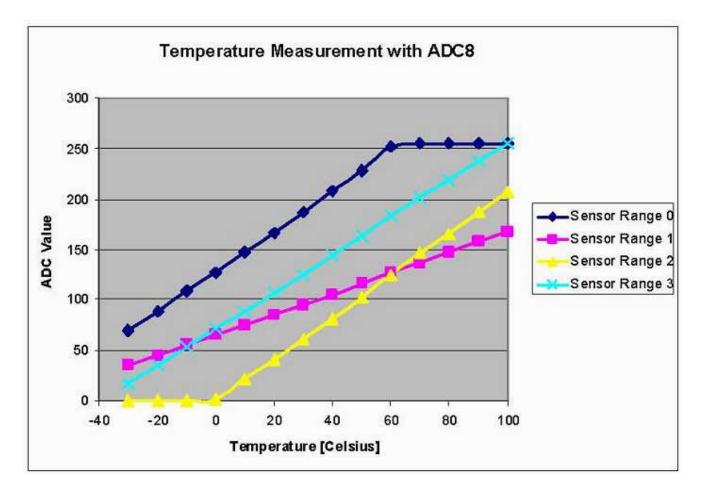


Figure 18. Temperature Ranges using ADC8





7.5. Low Battery Detector

A low battery detector (LBD) with digital read-out is integrated into the function is enabled by setting. A digital threshold may be programmed into the lbdt[4:0] field in "Register 1Ah. Low Battery Detector Threshold". When the digitized battery voltage reaches this threshold an interrupt will be generated on the nIRQ pin to the microcontroller. The microcontroller will then need to verify the interrupt by reading "Register 03h. Interrupt/Status 1" and "Register 04h. Interrupt/Status 2,". If the LBD is enabled while the module is in SLEEP mode, it will automatically enable the RC oscillator which will periodically turn on the LBD circuit to measure the battery voltage. The battery voltage may also be read out through "Register 1Bh. Battery Voltage Level" at any time when the LBD is enabled. The Low Battery Detect function is enabled by setting enlbd=1 in "Register 07h. Operating Mode and Function Control 1".

Ad	R/W	Function/Descri ption	D7	D6	D5	D4	D3	D2	D1	D0	POR Def.
1A	R/W	Low Battery Detector Threshold				lbdt[4]	lbdt[3]	lbdt[2]	lbdt[1]	lbdt[0]	14h
1B	R	Battery Voltage Level	0	0	0	vbat[4]	vbat[3]	vbat[2]	vbat[1]	vbat[0]	

The LBD output is digitized by a 5-bit ADC. When the LBD function is enabled, enlbd = 1 in "Register 07h. Operating Mode and Function Control 1", the battery voltage may be read at anytime by reading "Register 1Bh. Battery Voltage Level". A Battery Voltage Threshold may be programmed to register 1Ah. When the battery voltage level drops below the battery voltage threshold an interrupt will be generated on nIRQ pin to the microcontroller if the LBD interrupt is enabled in "Register 06h. Interrupt Enable 2,". The microcontroller will then need to verify the interrupt by reading the interrupt status register, Addresses 03 and 04H. The LSB step size for the LBD ADC is 50 mV, with the ADC range demonstrated in the table below. If the LBD is enabled the LBD and ADC will automatically be enabled every 1 s for approximately 250 µs to measure the voltage which minimizes the current consumption in Sensor mode. Before an interrupt is activated four consecutive readings are required.

BatteryVoltageV = 1.7+50*mV* X *ADCValue*

ADC Value	VDD Voltage [V]
0	< 1.7
1	1.7–1.75
2	1.75–1.8
29	3.1–3.15
30	3.15–3.2
31	>3.2



7.6. Wake-Up Timer

The module contains an integrated wake-up timer which periodically wakes the module from SLEEP mode. The wake-up timer runs from the internal 32.768 kHz RC Oscillator. The wake-up timer can be configured to run when in SLEEP mode. If enwt = 1 in "Register 07h. Operating Mode and Function Control 1" when entering SLEEP mode, the wake-up timer will count for a time specified by the Wake-Up Timer Period in Registers 10h-12h. At the expiration of this period an interrupt will be generated on the nIRQ pin if this interrupt is enabled. The microcontroller will then need to verify the interrupt by reading the Interrupt Status Registers 03h-04h. The wake-up timer value may be read at any time by the wtv[15:0] read only registers 13h-14h.

The formula for calculating the Wake-Up Period is the following:

W/UT =	$\frac{1 \times 2}{2.768} \times \frac{R \cdot D}{ms}$
WUT Register	Description
wtr[3:0]	R Value in Formula
wtd[1:0]	D Value in Formula
wtm[15:0]	M Value in Formula

Use of the D variable in the formula is only necessary if finer resolution is required than the R value gives.

Ad	R/W	Function/Descri ption	D7	D6	D5	D4	D3	D2	D1	D0	POR Def.
14	R/W	Wake-Up Timer Period 1			wtr[3]	wtr[2]	wtr[1]	wtr[0]	wtd[1]	wtd[0]	00h
15	R/W	Wake-Up Timer Period 2	wtm [15]	wtm[14]	wtm[13]	wtm[12]	wtm[11]	wtm[10]	wtm[9]	wtm[8]	00h
16	R/W	Wake-Up Timer Period 3	wtm [7]	wtm[6]	wtm[5]	wtm[4]	wtm[3]	wtm[2]	wtm[1]	wtm[0]	00h
17	R	Low Battery Detector Threshold	wtv[15]	wtv[14]	wtv[13]	wtv[12]	wtv[11]	wtv[10]	wtv[9]	wtv[8]	
18	R	Battery Voltage Level	wtv[7]	wt∨[6]	wtv[5]	wtv[4]	wtv[3]	wtv[2]	wtv[1]	wtv[0]	

There are two different methods for utilizing the wake-up timer (WUT) depending on if the WUT interrupt is enabled in "Register 06h. Interrupt Enable 2,". If the WUT interrupt is enabled then nIRQ pin will go low when the timer expires. The module will also change state so that the 30 M XTAL is enabled so that the microcontroller clock output is available for the microcontroller to use process the interrupt. The other method of use is to not enable the WUT interrupt and use the WUT GPIO setting. In this mode of operation the module will not change state until commanded by the microcontroller. The two different modes of operation of the WUT are demonstrated in Figure 19. A 32 kHz XTAL may also be used for better timing accuracy. By setting the x32 ksel bit in 07h, GPIO0 is automatically

reconfigured so that an external 32 kHz XTAL may be connected to this pin. In this mode, the GPIO0 is extremely sensitive to parasitic capacitance, so only the XTAL should be connected to this pin and the XTAL should be physically located as close to the pin as possible. Once the x32 ksel bit is set, all internal functions such as WUT, micro-controller clock, and LDC mode will use the 32 K XTAL and not the 32 kHz RC oscillator.



Interrupt Enable enwut=1 (Reg 06h)

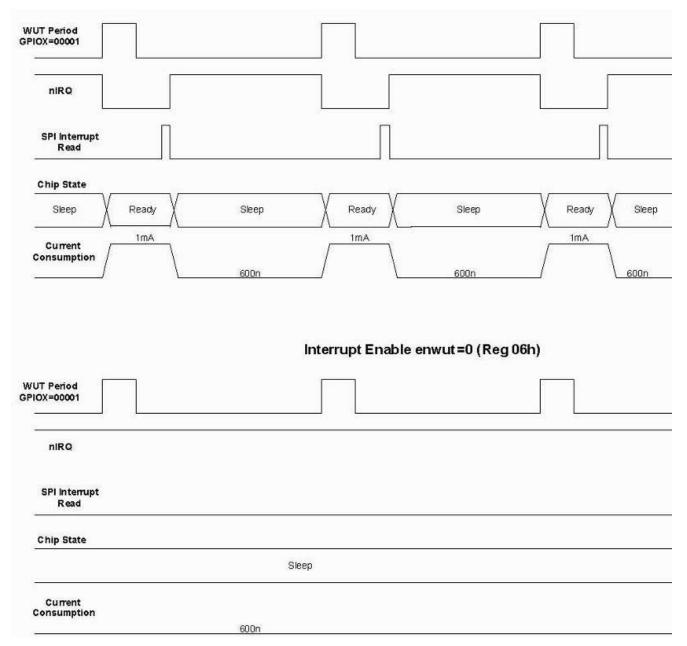


Figure 19. WUT Interrupt and WUT Operation

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7.7. GPIO Configuration

Three general purpose IOs (GPIOs) are available. Numerous functions such as specific interrupts, TRSW control, Microcontroller Output, etc. can be routed to the GPIO pins as shown in the tables below. When in Shutdown mode all the GPIO pads are pulled low.

Note: The ADC should not be selected as an input to the GPIO in Standby or Sleep Modes and will cause excess current consumption.

Add	R/W	Function/D escription	D7	D6	D5	D4	D3	D2	D1	D0	POR Def.
0B	R/W	GPIO0 Configuration	gpio0 drv[1]	gpio0dr v[1]	pup0	gpio0[4]	gpio0[3]	gpio0[2]	gpio0[1]	gpio0[0]	00h
0C	R/W	GPIO1 Configuration	Gpio1 drv[1]	wt gpio1dr v[1]	Pup1	gpio1[4]	gpio1[3]	gpio1[2]	gpio1[1]	gpio1[0]	00h
0D	R/W	GPIO2 Configuration	Gpio2 drv[1]	w gpio2dr v[1]	Pup2	gpio2[4]	gpio2[3]	gpio2[2]	gpio2[1]	gpio2[0]	00h
0E	R/W	I/O Port Configuration		extitst[2]	extitst[1]	extitst[0]	itsdo	dio2	dio1	dio0	00h

The GPIO settings for GPIO1 and GPIO2 are the same as for GPIO0 with the exception of the 00000 default setting. The default settings for each GPIO are listed below:

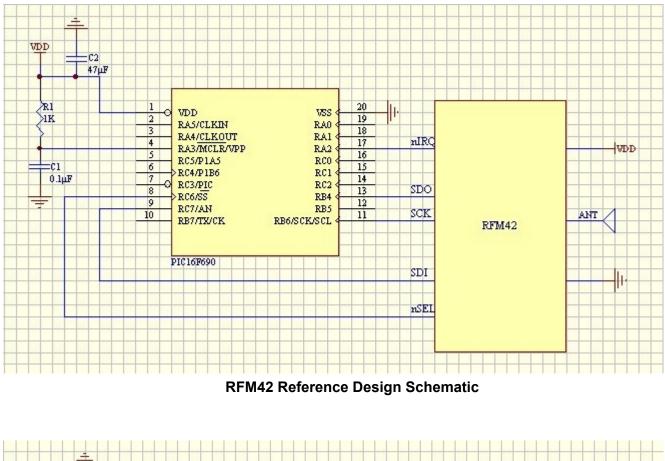
GPIO	00000—Default Setting
GPIO0	POR
GPIO1	POR Inverted
GPIO2	Microcontroller Clock

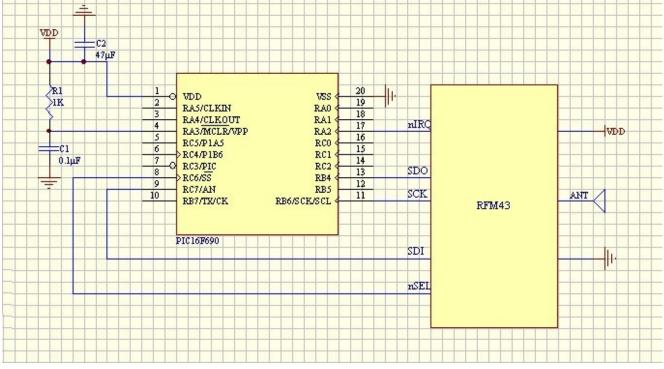
The module is configured to provide the System Clock output to the microcontroller so that only one crystal is needed in the system, therefore reducing the BOM cost. For the TX Data Source, Direct Mode is used because long packets are desired with a unique packet handling format already implemented in the microcontroller. In this configuration the TX Data Clock is configured onto GPIO0, the TX Data is configured onto GPIO1, and the Microcontroller System Clock output is configured onto GPIO2.

For a complete list of the available GPIO's see "Register 0Ch. GPIO Configuration 1,", "Register 0Dh. GPIO Configuration 2,", and "Register 0Eh. I/O Port Configuration,".



8. Reference Design

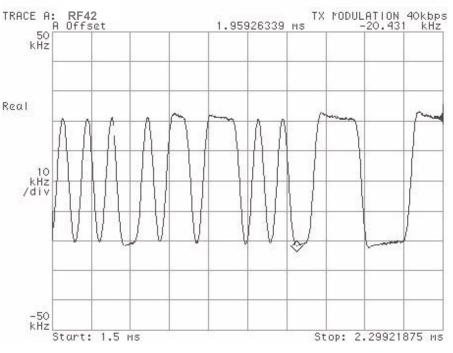




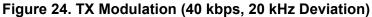


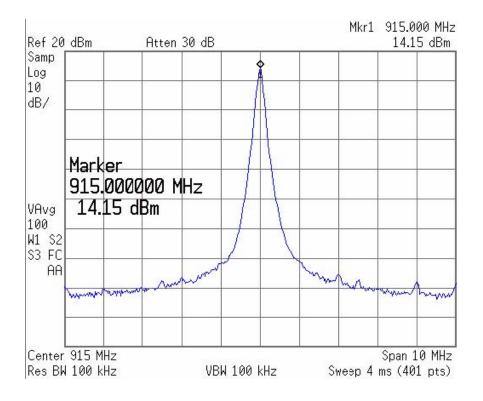


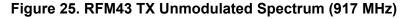
9. Measurement Results



Date: 04-22-08 Time: 02:28 PM









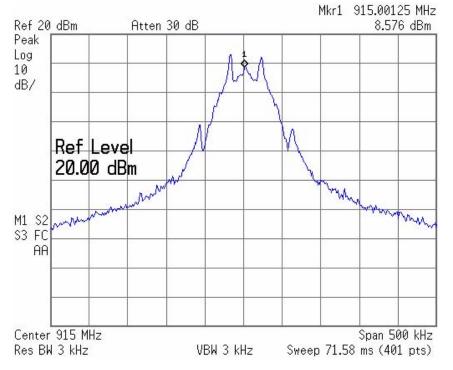
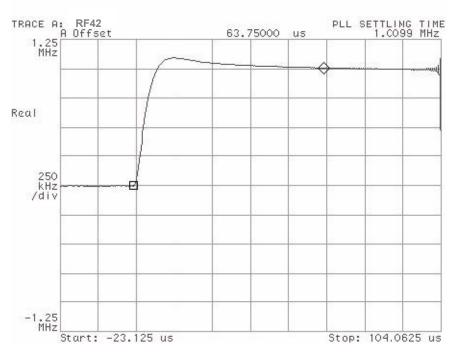


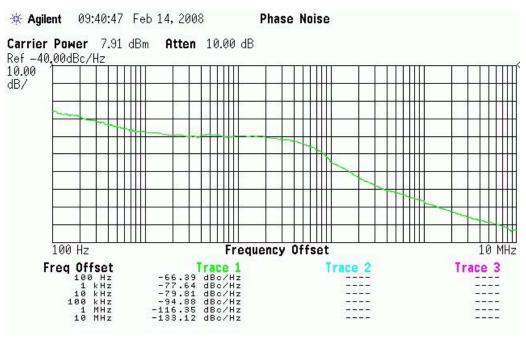
Figure 26. RFM43 TX Modulated Spectrum (917 MHz, 40 kbps, 20 kHz Deviation, GFSK)



Date: 04-23-08 Time: 04:03 PM

Figure 27. Synthesizer Settling Time for 1 MHz Jump Settled within 10 kHz









10. Reference Material

10.1. Complete Register Table and Descriptions Table 17. Register Descriptions

Add	R/W	Function/Desc				Data	1				POR
			D7	D6	D5	D4	D3	D2	D1	D0	Default
00	R	Device Type	0	0	0	dt[4]	dt[3]	dt[2]	dt[1]	dt[0]	
01	R	Device Version	0	0	0	vc[4]	vc[3]	vc[2]	vc[1]	vc[0]	04h
02	R	Device Status	ffovfl	ffunfl		headerr	reserved	reserved	cps[1]	cps[0]	_
03	R	Interrupt Status 1	ifferr	itxffafull	itxffaem	Reserved	iext	ipksent	Reserved	Reserved	_
04	R	Interrupt Status 2	iswdet	ipreaval	ipreainval	irssi	iwut	ilbd	ichiprdy	ipor	_
05	R/W	Interrupt Enable 1	enfferr	entxffafull	entxffaem	enrxffafull	enext	enpksent	enpkvalid	encrcerror	00h
06	R/W	Interrupt Enable 2	enswdet	enpreaval	enpreainval	enrssi	enwut	enlbd	enchiprdy	enpor	03h
07	R/W	Operating & Function Control 1	swres	enlbd	enwt	x32ksel	txon	rxon	pllon	xton	01h
08	R/W	Operating & Function Control 2	antdiv[2]	antdiv[1]	antdiv[0]	rxmpk	autotx	enidm	ffclrrx	ffclrtx	00h
09	R/W	Crystal Oscillator Load Capacitance	xtalshft	xlc[6]	xlc[5]	xlc[4]	xlc[3]	xlc[2]	xlc[1]	xlc[0]	7Fh
0A	R/W	Microcontroller Output Clock	Reserved	Reserved	clkt[1]	clkt[0]	enlfc	mclk[2]	mclk[1]	mclk[0]	06h
OB	R/W	GPIOD Configuration	gpio0drv[1]	gpio0drv[0]			gpio0[3]		gpio0[1]	gpio0[0]	00h
0D 0C	R/W				pup0	gpio0[4]		gpio0[2]			00h
0C 0D	R/W	GPIO1 Configuration	gpio1drv[1]	gpio1drv[0]	pup1	gpio1[4]	gpio1[3]	gpio1[2]	gpio1[1]	gpio1[0]	00h
		GPIO2 Configuration	gpio2drv[1]	gpio2drv[0]	pup2	gpio2[4]	gpio2[3]	gpio2[2]	gpio2[1]	gpio2[0]	
0E	R/W	I/O Port Configuration	Reserved	extitst[2]	extitst[1]	extitst[0]	itsdo	dio2	dio1	dio0	00h
OF	R/W	ADC Configuration	adcstart/adc- done	adcsel[2]	adcsel[1]	adcsel[0]	adcref[1]	adcref[0]	adcgain[1]	adcgain[0]	00h
10	R/W	ADC Sensor Amplifier Offset	Reserved	Reserved	Reserved	Reserved	adcoffs[3]	adcoffs[2]	adcoffs[1]	adcoffs[0]	00h
11	R	ADC Value	adc[7]	adc[6]	adc[5]	adc[4]	adc[3]	adc[2]	adc[1]	adc[0]	_
12	R/W	Temperature Sensor Control	tsrange[1]	tsrange[0]	entsoffs	entstrim	tstrim[3]	tstrim[2]	tstrim[1]	tstrim[0]	20h
13	R/W	Temperature Value Offset	tvoffs[7]	tvoffs[6]	tvoffs[5]	tvoffs[4]	tvoffs[3]	tvoffs[2]	tvoffs[1]	tvoffs[0]	00h
14	R/W	Wake-Up Timer Period 1	Reserved	Reserved	Reserved	wtr[4]	wtr[3]	wtr[2]	wtr[1]	wtr[0]	03h
15	R/W	Wake-Up Timer Period 2	wtm[15]	wtm[14]	wtm[13]	wtm[12]	wtm[11]	wtm[10]	wtm[9]	wtm[8]	00h
16	R/W	Wake-Up Timer Period 3	wtm[7]	wtm[6]	wtm[5]	wtm[4]	wtm[3]	wtm[2]	wtm[1]	wtm[0]	01h
17	R	Wake-Up Timer Value 1	wtv[15]	wtv[14]	wtv[13]	wtv[12]	wtv[11]	wtv[10]	wtv[9]	wtv[8]	
18	R	Wake-Up Timer Value 2	wtv[7]	wtv[6]	wtv[5]	wtv[4]	wtv[3]	wtv[2]	wtv[1]	wtv[0]	
19	RW	Low-Duty Cycle Mode Duration	Idc[7]	Idc[6]	Idc[5]	Idc[4]	Idc[3]	Idc[2]	Idc[1]	Idc[0]	00h
19 1A	R/W		• •	• •							14h
		Low Battery Detector Threshold	Reserved	Reserved	Reserved	lbdt[4]	lbdt[3]	lbdt[2]	lbdt[1]	lbdt[0]	14N
1B	R	Battery Voltage Level	0	0	0	vbat[4]	vbat[3]	vbat[2]	vbat[1]	vbat[0]	_
1C-2F					Reserved						
30	R/W	Data Access Control	enpacrx	Isbfrst	credonly	Reserved	enpactx	encrc	crc[1]	crc[0]	1Dh
31	R	EzMAC status	0	rxcrc1	pksrch	pkrx	pkvalid	crcerror	pktx	pksent	_
32					Reserved						
33	R/W	Header Control 2	Reserved	hdlen[2]	hdlen[1]	hdlen[0]	fixpklen	synclen[1]	synclen[0]	prealen[8]	22h
34	R/W	Preamble Length	prealen[7]	prealen[6]	prealen[5]	prealen[4]	prealen[3]	prealen[2]	prealen[1]	prealen[0]	08h
36	R/W	Sync Word 3	sync[31]	sync[30]	sync[29]	sync[28]	sync[27]	sync[26]	sync[25]	sync[24]	2Dh
37	R/W	Sync Word 2	sync[23]	sync[22]	sync[21]	sync[20]	sync[19]	sync[18]	sync[17]	sync[16]	D4h
38	R/W	Sync Word 1	sync[15]	sync[14]	sync[13]	sync[12]	sync[11]	sync[10]	sync[9]	sync[8]	00h
39	R/W	Sync Word 0	sync[7]	sync[6]	sync[5]	sync[4]	sync[3]	sync[2]	sync[1]	sync[0]	00h
3A	R/W	Transmit Header 3	txhd[31]	txhd[30]	txhd[29]	txhd[28]	txhd[27]	txhd[26]	txhd[25]	txhd[24]	00h
3B	R/W	Transmit Header 2	txhd[23]	txhd[22]	txhd[21]	txhd[20]	txhd[19]	txhd[18]	txhd[17]	txhd[16]	00h
3C	R/W	Transmit Header 1	txhd[15]	txhd[14]	txhd[13]	txhd[12]	txhd[11]	txhd[10]	txhd[9]	txhd[8]	00h
3D	R/W	Transmit Header 0	txhd[7]	txhd[6]	txhd[5]	txhd[4]	txhd[3]	txhd[2]	txhd[1]	txhd[0]	00h
3E	R/W	Transmit Packet Length	pklen[7]	pklen[6]	pklen[5]	pklen[4]	pklen[3]	pklen[2]	pklen[1]	pklen[0]	00h
3F-50		Alanamic Packet Congel	phon[1]	phicitol	Reserved	phien[4]	phicital	phion[2]	prioriti	phicitol	0011
51	R/W	Digital Test Bus	Reserved	ensctest	dtb[5]	dtb[4]	d#b/21	dtb[2]	dtb[1]	dtb[0]	00h
51	R/W	2					dtb[3]		• •		2Fh
		TX Ramp Control	Reserved	txmod[2]	txmod[1]	txmod[0]	Idoramp[1]	Idoramp[0]	txramp[1]	txramp[0]	
53 54	R/W R/W	PLL Tune Time Invalid Preamble Threshold and PA Misc	plits[4] Reserved	plits[3] Reserved	pllts[2] inv_pre_th[3]	pllts[1] inv_pre_th[2]	pllts[0] inv_pre_th[1]	pllt0[2] inv_pre_th[0]	plit0[1] Ido_pa_boost	pllt0[0] pa_vbias_ boost	52h 14h
55	R/W	Calibration Control	Reserved	xtalstarthalf	Reserved	enrcfcal	rccal	vcocaldp	vcocal	skipyco	44h
56	R/W	Modem Test					OOkth				00h
			borfbyp	slicfbyp	dttype	oscdeten		refclksel	refclkinv	distogg	
57	R/W	Chargepump Test	pfdrst	fbdiv_rst	cpforceup	cpforcedn	cdconly	cdccur[2]	cdccur[1]	cdccur[0]	00h
58	R/W	Chargepump Current Trimming/Override	cpcurr[1]	cpcurr[0]	срсоптоу	cpcorr[4]	cpcorr[3]	cpcorr[2]	cpcorr[1]	cpcorr[0]	80h
59	R/W	Divider Current Trimming	txcorboosten	fbdivhc	d3trim[1]	d3trim[0]	d2trim[1]	d2trim[0]	d1p5trim[1]	d1p5trim[0]	40h
5A	R/W	VCO Current Trimming	txcurboosten	vcocorrov	vcocorr[3]	vcocorr[2]	vcocorr[1]	vcocorr[0]	vcocur[1]	vcocur[0]	03h
5B	R/W	VCO Calibration / Override	vcocalov/vcdone	vcocal[6]	vcocal[5]	vcocal[4]	vcocal[3]	vcocal[2]	vcocal[1]	vcocal[0]	00h
5C	R/W	Synthesizer Test	dsmdt	vcotype	enoloop	dsmod	dsorder[1]	dsorder[0]	dsrstmod	dsrst	0Eh
5D	R/W	Block Enable Override 1	enmix	enina	enpga	enpa	enbf5	endv32	enbf12	enmx2	00h
5E	R/W	Block Enable Override 2	ends	enidet	enmx3	enbf4	enbf3	enbf11	enbf2	pllreset	40h
5F	R/W	Block Enable Override 3	enfrdv	endv31	endv2	endv1p5	dvbshunt	envco	encp	enbg	00h
60	R/W	Channel Filter Coefficient Address	Reserved	Reserved	Reserved	Reserved	chfiladd[3]	chfiladd[2]	chfiladd[1]	chfiladd[0]	00h
61	R/W	Channel Filter Coefficient Value	Reserved	Reserved	chfilval[5]	chfilval[4]	chfilval[3]	chfilval[2]	chfilval[1]	chfilval[0]	00h
01	1000	Sharmer mer Odellident value	110301VCU	neachied	crimedi[0]	Grinten[4]	crimento]	sinnvai[2]	crimvei[1]	crimval[0]	0011



Table 17. Register Descriptions (Continued)

Add	R/W	Function/Desc				Data	1				POR
			D7	D6	D5	D4	D3	D2	D1	D0	Default
62	R/W	Crystal Oscillator / Control Test	pwst[2]	pwst[1]	pwst[0]	clkhyst	enbias2x	enamp2x	bufovr	enbuf	24h
63	R/W	RC Oscillator Coarse Calibration/Override	rccov	rcc[6]	rcc[5]	rcc[4]	rcc[3]	rcc[2]	rcc[1]	rcc[0]	00h
64	R/W	RC Oscillator Fine Calibration/Override	rcfov	rcf[6]	rcf[5]	rcf[4]	rcf[3]	rcf[2]	rcf[1]	rcf[0]	00h
65	R/W	LDO Control Override	enspor	enbias	envcoldo	enifldo	enrfldo	enpllido	endigldo	endigpwdn	81h
66	R/W	LDO Level Setting	enovr	enxtal	ents	enrc32	Reserved	diglvl[2]	diglvl[1]	diglvl[0]	02h
67-6A					Reserved			-			
6B	R/W	GFSK FIR Filter Coefficient Address	Reserved	Reserved	Reserved	Reserved	Reserved	firadd[2]	firadd[1]	firadd[0]	00h
6C	R/W	GFSK FIR Filter Coefficient Value	Reserved	Reserved	firval[5]	firval[4]	firval[3]	firval[2]	firval[1]	firval[0]	01h
6D	R/W	TX Power	Reserved	Reserved	Reserved	Reserved	Ina_sw	txpow[2]	txpow[1]	txpow[0]	08h
6E	R/W	TX Data Rate 1	txdr[15]	txdr[14]	txdr[13]	txdr[12]	txdr[11]	txdr[10]	txdr[9]	txdr[8]	0Ah
6F	R/W	TX Data Rate 0	txdr[7]	txdr[6]	txdr[5]	txdr[4]	txdr[3]	txdr[2]	txdr[1]	txdr[0]	3Dh
70	R/W	Modulation Mode Control 1	Reserved	Reserved	txdtrtscale	enphpwdn	manppol	enmaninv	enmanch	enwhite	0Ch
71	R/W	Modulation Mode Control 2	trclk[1]	trclk[0]	dtmod[1]	dtmod[0]	eninv	fd[8]	modtyp[1]	modtyp[0]	00h
72	R/W	Frequency Deviation	fd[7]	fd[6]	fd[5]	fd[4]	fd[3]	fd[2]	fd[1]	fd[0]	20h
73	R/W	Frequency Offset 1	fo[7]	fo[6]	fo[5]	fo[4]	fo[3]	fo[2]	fo[1]	fo[0]	00h
74	R/W	Frequency Offset 2	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	fo[9]	fo[8]	00h
75	R/W	Frequency Band Select	Reserved	sbsel	hbsel	fb[4]	fb[3]	fb[2]	fb[1]	fb[0]	75h
76	R/W	Nominal Carrier Frequency 1	fc[15]	fc[14]	fc[13]	fc[12]	fc[11]	fc[10]	fc[9]	fc[8]	BBh
77	R/W	Nominal Carrier Frequency 0	fc[7]	fc[6]	fc[5]	fc[4]	fc[3]	fc[2]	fc[1]	fc[0]	80h
78	R/W	Miscellaneous Settings	Reserved	Reserved	Reserved	Reserved	Alt_PA_Seq	rcosc[2]	rcosc[1]	rcosc[0]	09h
79	R/W	Frequency Hopping Channel Select	fhch[7]	fhch[6]	fhch[5]	fhch[4]	fhch[3]	fhch[2]	fhch[1]	fhch[0]	00h
7A	R/W	Frequency Hopping Step Size	fhs[7]	fhs[6]	fhs[5]	fhs[4]	fhs[3]	fhs[2]	fhs[1]	fhs[0]	00h
7B	R/W	Turn Around and 15.4 Length Compliance	15.4 Length	Reserved	Reserved	Reserved	Reserved	turn_around _en	Phase[1]	Phase[0]	09h
7C	R/W	TX FIFO Control 1	Reserved	Reserved	txafthr[5]	txafthr[4]	txafthr[3]	txafthr[2]	txafthr[1]	txafthr[0]	37h
7D	R/W	TX FIFO Control 2	Reserved	Reserved	txaethr[5]	txaethr[4]	txaethr[3]	txaethr[2]	txaethr[1]	txaethr[0]	04h
7E					Reserved						
7F	R/W	FIFO Access	fifod[7]	fifod[6]	fifod[5]	fifod[4]	fifod[3]	fifod[2]	fifod[1]	fifod[0]	_



Register 00h. Device Type Code (DT)

Bit	D7	D6	D5	D4	D3	D2	D1	D0			
Name		Reserved		dt[4:0]							
Туре			R			R					

Reset value = 00000111

Bit	Name	Function
7:5	Reserved	Reserved.
4:0	dt[4:0]	Device Type Code.

Register 01h. Version Code (VC)

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	Reserved			VC[4:0]				
Туре	R					R		

Reset value = xxxxxxxx

Bit	Name	Function
7:5	Reserved	Reserved.
		Version Code.
4.0	V0[4:0]	Code indicating the version of the module.
4:0	VC[4:0]	Rev A0: 00100
		Rev V2: 00011



Register 02h. Device Status

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	ffovfl	ffunfl	Reserved	Reserved	Reserved	Reserved	cps[1:0]	ffovfl
Туре	R	R	R	R	R	R	R	R

Reset value = xxxxxxxx

Bit	Name	Function
7	ffovfl	тх
6	ffunfl	тх
5:4	Reserved	Reserved.
3:2	Reserved	Reserved.
1:0	cps[1:0]	Module Power State.
		00: Idle State
		10: TX State



Register 03h. Interrupt/Status 1

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	ifferr	itxffafull	ixtffaem	Reserved	iext	ipksent	Reserved	Reserved
Туре	R	R	R	R	R	R	R	R

Reset value = xxxxxxxx

Bit	Name	Function
7	ifferr	FIFO Underflow/Overflow Error.
1		When set to 1 the TX FIFO has overflowed or underflowed.
		TX FIFO Almost Full.
6	itxffafull	When set to 1 the TX FIFO has met its almost full threshold and needs to be
		transmitted.
5	itxffaem	TX FIFO Almost Empty.
5	itxilaeili	When set to 1 the TX FIFO is almost empty and needs to be filled.
4	Reserved	Reserved.
		External Interrupt.
3	iext	When set to 1 an interrupt occurred on one of the GPIO's if it is programmed so. The
		status can be checked in register 0Eh. See GPIOx Configuration section for the details.
2	ipksent	Packet Sent Interrupt.
2	ipksent	When set to1 a valid packet has been transmitted.
1:0	Reserved	Reserved.

When any of the Interrupt/Status 1 bits change state from 0 to 1 the device will notify the microcontroller by setting the nIRQ pin LOW if it is enabled in the Interrupt Enable 1 register. The nIRQ pin will go to HIGH and all the **enabled** interrupt bits will be cleared when the microcontroller reads this address. If any of these bits is not enabled in the Interrupt Enable 1 register then it becomes a status signal that can be read anytime in the same location and will not be cleared by reading the register.





Table 18. Interrupt or Status 1 Bit Set/Clear Description

Bit	Status	Set/Clear Conditions
Ы	Name	
7	ifferr	Set if there is a FIFO overflow or underflow. Cleared by applying FIFO reset.
		Set when the number of bytes written to TX FIFO is greater than the Almost Full
6	itxffafull	threshold. Automatically cleared at the start of transmission when the number of
		bytes in the FIFO is less than or equal to the threshold.
		Set when the number of bytes in the TX FIFO is less than or equal to the Almost
5	itxffaem	Empty threshold. Automatically cleared when the number of data bytes in the TX
		FIFO is above the Almost Empty threshold.
4	Reserved	Reserved.
3	iext	External interrupt source.
2	inkoont	Set once a packet is successfully sent (no TX abort). Cleared upon leaving FIFO
2	ipksent	mode or at the start of a new transmission.
1:0	Reserved	Reserved.

Table 19. When are Individual Status Bits Set/Cleared if not Enabled as Interrupts?

D:4	Status	Set/Clear Conditions
Bit	Name	
7	7 ifferr	Set if there is a FIFO Overflow or Underflow. It is cleared only by applying FIFO
1	men	reset to the specific FIFO that caused the condition.
		Will be set when the number of bytes written to TX FIFO is greater than the
6	itxffafull	Almost Full threshold set by SPI. It is automatically cleared when we start
0	itxilaiuli	transmitting and the FIFO data is read out and the number of bytes left in the
		FIFO is smaller or equal to the threshold).
		Will be set when the number of bytes (not yet transmitted) in TX FIFO is smaller
5	itxffaem	or equal than the Almost Empty threshold set by SPI. It is automatically cleared
5		when we write enough data to TX FIFO so that the number of data bytes not yet
		transmitted is above the Almost Empty threshold.
4	Reserved	Reserved.
3	iext	External interrupt source.
		Will go high once a packet is sent all the way through (no TX abort). This status
2	ipksent	will be cleaned if 1) We leave FIFO mode or 2) In FIFO mode we start a new
		transmission.
1:0	Reserved	Reserved.



Register 04h. Interrupt/Status 2

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	Reserved			iwut	ilbd	ichiprdy	ipor	
Туре	R			R	R	R	R	

Reset value = xxxxxxxx

Bit	Name	Function
7:4	Reserved	Reserved.
3	iwut	Wake-Up-Timer.
3	iwut	On the expiration of programmed wake-up timer this bit will be set to 1.
		Low Battery Detect.
2	ilbd	When a low battery event is been detected this bit will be set to 1. This interrupt
2	libu	event is saved even if it is not enabled by the mask register bit and causes an
		interrupt after it is enabled.
1	ichiprdy	Module Ready (XTAL).
I	ichiprdy	When a module ready event has been detected this bit will be set to 1.
		Power-on-Reset (POR).
0	ipor	When the module detects a Power on Reset above the desired setting this bit will
		be set to 1.

When any of the Interrupt/Status Register 2 bits change state from 0 to 1 the control block will notify the microcontroller by setting the nIRQ pin LOW if it is enabled in the Interrupt Enable 2 register. The nIRQ pin will go to HIGH and all the **enabled** interrupt bits will be cleared when the microcontroller reads this address. If any of these bits is not enabled in the Interrupt Enable 2 register then it becomes a status signal that can be read anytime in the same location and will not be cleared by reading the register.



Table 20. Interrupt or Status 2 Bit Set/Clear Description

Bit	Name	Set/Clear Conditions
7: 4	Reserved	Reserved.
3	iwut	Wake time timer interrupt. Use as an interrupt, not as a status.
2	ilbd	Low Battery Detect. When a low battery event is been detected this bit will be set to 1. This interrupt event is saved even if it is not enabled by the mask register bit and causes an interrupt after it is enabled. Probably the status is cleared once the battery is replaced.
1	ichiprdy	Module ready goes high once we enable the xtal, TX and a settling time for the Xtal clock elapses. The status stay high unless we go back to Idle mode.
0	ipor	Power on status.

Table 21. Detailed Description of Status Registers when not Enabled as Interrupts

Bit	Name	Set/Clear Conditions
7:4	Reserved	Reserved.
3	iwut	Wake time timer interrupt. Use as an interrupt, not as a status.
2	ilbd	Low Battery Detect. When a low battery event is been detected this bit will be set to 1. This interrupt event is saved even if it is not enabled by the mask register bit and causes an interrupt after it is enabled. Probably the status is cleared once the battery is replaced.
1	ichiprdy	Module ready goes high once we enable the xtal, TX and a settling time for the Xtal clock elapses. The status stay high unless we go back to Idle mode.
0	ipor	Power on status.



Register 05h. Interrupt Enable 1

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	enfferr	entxffafull	entxffaem	Reserved	enext	enpksent	Reserved	Reserved
Туре	R/w	R/w	R/w	R/w	R/w	R/w	R/w	R/w

Bit	Name	Function		
7	enfferr	Enable FIFO Underflow/Overflow.		
1	ennen	When set to 1 the FIFO Underflow/Overflow interrupt will be enabled.		
6	entxffafull	Enable TX FIFO Almost Full.		
0	o entxitatuli	When set to 1 the TX FIFO Almost Full interrupt will be enabled.		
5	entxffaem	Enable TX FIFO Almost Empty.		
5	entxilaem	When set to 1 the TX FIFO Almost Empty interrupt will be enabled.		
4	Reserved	Reserved.		
3	enext	Enable External Interrupt.		
3	enext	When set to 1 the External Interrupt will be enabled.		
2	oppkoopt	Enable Packet Sent.		
2	enpksent	When ipksent =1 the Packet Sense Interrupt will be enabled.		
1:0	Reserved	Reserved.		



Register 06h. Interrupt Enable 2

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name		Rese	erved		enwut	enlbd	enchiprdy	enpor
Туре		R				R/w	R/w	R/w

Bit	Name	Function				
7:4	Reserved	Reserved.				
3	enwut	Enable Wake-Up Timer.				
3	enwut	When set to 1 the Wake-Up Timer interrupt will be enabled.				
2	enlbd	Enable Low Battery Detect.				
2	enibu	When set to 1 the Low Battery Detect interrupt will be enabled.				
1	enchiprdy	Enable Module Ready (XTAL).				
l	enchipidy	When set to 1 the Module Ready interrupt will be enabled.				
0	oppor	Enable POR.				
0	enpor	When set to 1 the POR interrupt will be enabled.				



Register 07h. Operating Mode and Function Control 1

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	swres	enlbd	enwt	x32ksel	txon	Reserved	pllon	xton
Туре	R/w	R/w	R/w	R/w	R/w	R/w	R/w	R/w

Bit	Name	Function
		Software Register Reset Bit.
7	swres	This bit may be used to reset all registers simultaneously to a DEFAULT state,
1	SWIES	without the need for sequentially writing to each individual register. The RESET is
		accomplished by setting swres = 1. This bit will be automatically cleared.
		Enable Low Battery Detect.
6	enlbd	When this bit is set to 1 the Low Battery Detector circuit and threshold
		comparison will be enabled.
		Enable Wake-Up-Timer.
5	enwt	Enabled when enwt = 1. If the Wake-up-Timer function is enabled it will operate in
5	Cliwi	any mode and notify the microcontroller through the GPIO interrupt when the timer
		expires.
		32,768 kHz Crystal Oscillator Select.
4	x32ksel	0: RC oscillator
		1: 32 kHz crystal
		TX on in Manual Transmit Mode.
		Automatically cleared in FIFO mode once the packet is sent. Transmission can be
3	txon	aborted during packet transmission, however, when no data has been sent yet,
		transmission can only be aborted after the device is programmed to "unmodulated
		carrier" ("Register 71h. Modulation Mode Control 2").
2	Reserved	Reserved.
		TUNE Mode (PLL is ON).
1	pllon	When pllon = 1 the PLL will remain enabled in Idle State. This will for faster
		turn-around time at the cost of increased current consumption in Idle State.
0	xton	READY Mode (Xtal is ON).



Register 08h. Operating Mode and Function Control 2

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name		Rese	erved		autotx	Rese	erved	ffclrtx
Туре		R/w				R/w	R/w	R/w

Reset value = 00000000

Bit	Name	Function
7:4	Reserved	Reserved.
		Automatic Transmission.
3	autotx	When autotx = 1 the transceiver will enter automatically TX State when the
3	autotx	FIFO is almost full. When the FIFO is empty it will automatically return to the
		Idle State.
2: 1	Reserved	Reserved.
		TX FIFO Reset/Clear.
0	ffclrtx	This has to be a two writes operation: Setting ffclrtx =1 followed by ffclrtx = 0
		will clear the contents of the TX FIFO.

Register 09h. 30 MHz Crystal Oscillator Load Capacitance

Bit	D7	D6	D5	D4	D3	D2	D1	D0		
Name	xtalshft		xlc[6:0]							
Туре	R/w				R/w					

Bit	Name	Function
		Additional capacitance to course shift the frequency if xlc[6:0] is not sufficient.
7	xtalshft	Not binary
		with xlc[6:0].
6:0	xlc[6:0]	Tuning Capacitance for the 30 MHz XTAL.



Register 0Ah. Microcontroller Output Clock

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	Res	Reserved		clkt[1:0]		mclk[2:0]		
Туре	R	1	R/	R/w R/w		R/w		

Reset value = xx000110

Bit	Name	Function
7:6	Reserved	Reserved.
5:4	clkt[1:0]	Clock Tail.If enlfc = 0 then it can be useful to provide a few extra cycles for the microcontroller to complete its operation. Setting the clkt[1:0] register will provide the addition cycles of the clock before it shuts off.00:0 cycle01:128 cycles10:256 cycles11:512 cycles
3	enlfc	Enable Low Frequency Clock. When enlfc = 1 and the module is in Sleep mode then the 32.768 kHz clock will be provided to the microcontroller no matter what the selection of mclk[2:0] is. For example if mclk[2:0] = '000', 30 MHz will be available through the GPIO to output to the microcontroller in all Idle or TX states. When the module is commanded to Sleep mode the 30 MHz clock will become 32.768 kHz.
2:0	mclk[2:0]	Microcontroller Clock.Different clock frequencies may be selected for configurable GPIO clockoutput. All clock frequencies are created by dividing the XTAL except for the 32kHz clock which comes directly from the 32 kHz RC Oscillator. The mclk[2:0]setting is only valid when xton = 1 except the 111.000:30 MHz001:15 MHz010:10 MHz011:4 MHz100:3 MHz101:2 MHz110:1 MHz111:32.768 kHz



Register 0Bh. GPIO Configuration 0

Bit	D7	D6	D5	D4	D3	D2	D1	D0	
Name	gpiodrv0[1:0]		pup0	gpio0[4:0]					
Туре	R/w		R/w	R/w					

GPIO Driving Capability Setting. Pullup Resistor Enable on GPIO0.						
internally between VDD and						
ed (output)						
elow threshold setting (output)						
nction with TX Data pin						
.1)						
ut)						
)						



Register 0Ch. GPIO Configuration 1

Bit	D7	D6	D5	D4	D3	D2	D1	D0	
Name	gpiodrv1[1:0]		pup1	Gpio1[4:0]					
Туре	R/w		R/w	R/w					

Bit	Name	Function						
7:6	gpiodrv1[1:0]	GPIO Driving Capability Setting.						
		Pullup Resistor Enable on GPIO1.						
5	Pup1	When set to 1 the a 200 k $_{\rm c}$ resistor is connected internally between VDD and						
		the pin if the GPIO is configured as a digital input.						
		GPIO1 pin Function Select.						
		00000: Inverted Power-On-Reset (output)						
		00001: Wake-Up Timer: 1 when WUT has expired (output)						
		00010: Low Battery Detect: 1 when battery is below threshold setting (output)						
		00011: Direct Digital Input						
		00100: External Interrupt, falling edge (input)						
		00101: External Interrupt, rising edge (input)						
		00110: External Interrupt, state change (input)						
		00111: ADC Analog Input						
		01000: Reserved (Analog Test N Input)						
		01001: Reserved (Analog Test P Input)						
		01010: Direct Digital Output						
		01011: Reserved (Digital Test Output)						
		01100: Reserved (Analog Test N Output)						
		01101: Reserved (Analog Test P Output)						
		01110: Reference Voltage (output)						
4:0	gpio1[4:0]	01111: TX Data CLK output to be used in conjunction with TX Data pin						
		(output)						
		10000: TX Data input for direct modulation (input)						
		10001: External Retransmission Request (input)						
		10010: TX State (output)						
		10011: TX FIFO Almost Full (output) 10100: Reserved						
		10100. Reserved						
		10110: Reserved						
		10111: Reserved						
		11000: Reserved						
		11001: Reserved						
		11010: Reserved						
		11011: Reserved						
		11100: Reserved						
		11101: VDD						
		else : GND						



Register 0Ch. GPIO Configuration 2

Bit	D7	D6	D5	D4	D3	D2	D1	D0	
Name	gpiodrv2[1:0]		pup2	Gpio2[4:0]					
Туре	R/\	N	R/w	R/w					

Bit	Name	Function
7:6	gpiodrv2[1:0]	GPIO Driving Capability Setting.
		Pullup Resistor Enable on GPIO2.
5	Pup2	When set to 1 the a 200 k $_{\rm .}$ resistor is connected internally between VDD and
		the pin if the GPIO is configured as a digital input.
		GPIO1 pin Function Select.
		00000: Inverted Power-On-Reset (output)
		00001: Wake-Up Timer: 1 when WUT has expired (output)
		00010: Low Battery Detect: 1 when battery is below threshold setting (output)
		00011: Direct Digital Input
		00100: External Interrupt, falling edge (input)
		00101: External Interrupt, rising edge (input)
		00110: External Interrupt, state change (input)
		00111: ADC Analog Input
		01000: Reserved (Analog Test N Input)
		01001: Reserved (Analog Test P Input)
		01010: Direct Digital Output
		01011: Reserved (Digital Test Output)
		01100: Reserved (Analog Test N Output)
		01101: Reserved (Analog Test P Output)
		01110: Reference Voltage (output)
4:0	gpio2[4:0]	01111: TX Data CLK output to be used in conjunction with TX Data pin
		(output)
		10000: TX Data input for direct modulation (input)
		10001: External Retransmission Request (input)
		10010: TX State (output)
		10011: TX FIFO Almost Full (output)
		10100: Reserved 10101: Reserved
		10101: Reserved 10110: Reserved
		10111: Reserved
		11000: Reserved
		11001: Reserved
		11010: Reserved
		11011: Reserved
		11100: Reserved
		11101: VDD
		else : GND
L		



Register 0Eh. I/O Port Configuration

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	Reserved	extitst[2]	extitst[1]	extitst[0]	itsdo	dio2	dio1	dio0
Туре	R	R	R	R	R/w	R/w	R/w	R/w

Bit	Name	Function
7	Reserved	Reserved
6	extitst[2]	External Interrupt Status. If the GPIO2 is programmed to be external interrupt sources then the status can be read here.
5	extitst[1]	External Interrupt Status. If the GPIO1 is programmed to be external interrupt sources then the status can be read here.
4	extitst[0]	External Interrupt Status. If the GPIO0 is programmed to be external interrupt sources then the status can be read here.
3	itsdo	Interrupt Request Output on the SDO Pin. nIRQ output is present on the SDO pin if this bit is set and the nSEL input is inactive (high).
2	dio2	Direct I/O for GPIO2. If the GPIO2 is configured to be a direct output then the value on the GPIO pin can be set here. If the GPIO2 is configured to be a direct input then the value of the pin can be read here.
1	dio1	Direct I/O for GPIO1. If the GPIO1 is configured to be a direct output then the value on the GPIO pin can be set here. If the GPIO1 is configured to be a direct input then the value of the pin can be read here.
0	dio0	Direct I/O for GPIO0. If the GPIO0 is configured to be a direct output then the value on the GPIO pin can be set here. If the GPIO0 is configured to be a direct input then the value of the pin can be read here.



Register 0Fh. ADC Configuration

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	adcstart/ adcdone	adcsel[2:0]			adcre	ef[1:0]	adcga	ain[1:0]
Туре	R	R	R	R	R/w	R/w	R/w	R/w

Bit	Name	Function
7	adcstart/adcdone	ADC Measurement Start Bit.
1		Reading this bit gives 1 if the ADC measurement cycle has been finished.
		ADC Input Source Selection. The internal 8-bit ADC input source can be selected as follows:
		000: Internal Temperature Sensor
		001: GPIO0, single-ended
6:4	odocol[2:0]	010: GPIO1, single-ended
0.4	adcsel[2:0]	011: GPIO2, single-ended
		100: GPIO0(+) – GPIO1(–), differential
		101: GPIO1(+) – GPIO2(–), differential
		110: GPIO0(+) – GPIO2(–), differential
		111: GND
		ADC Reference Voltage Selection.
		The reference voltage of the internal 8-bit ADC can be selected as follows:
3:2	adcref[1:0]	0X: bandgap voltage (1.2 V)
		10: VDD / 3
		11: VDD / 2
		ADC Sensor Amplifier Gain Selection.
		The full scale range of the internal 8-bit ADC in differential mode (see adcsel)
1:0	adcgain[1:0]	can be set
1.0	aucyanii[1.0]	as follows:
		adcref[0] = 0: adcref[0] = 1:
		FS = 0.014 x (adcgain[1:0] + 1) x VDD FS = 0.021 x (adcgain[1:0] + 1) x VDD



Register 10h. ADC Sensor Amplifier Offset

Bit	D7	D6	D5	D4	D3	D2	D1	D0	
Name		Res	erved		adcoffs[3:0]				
Туре		F	र			R/w	1		

Reset value = xxxx0000

Bit	Name	Function				
7:4	Reserved	Reserved.				
3: 0	adcoffs[3:0]	ADC Sensor Amplifier Offset*.				
*Note: The o	*Note: The offset can be calculated as Offset = adcoffs[2:0] x VDD / 1000; MSB = adcoffs[3] = Sign bit.					

Register 11h. ADC Value

Bit	D7	D6	D5	D4	D3	D2	D1	D0			
Name		adc[7:0]									
Туре		R									

Reset value = xxxxxxxx

Bit	Name	Function
7: 0	adc[7:0]	Internal 8 bit ADC Output Value.



Register 12h. Temperature Sensor Calibration

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	tsrange[1:0]		entsoffs	entstrim		tstrim[3:0]		
Туре	R/w		R/w	R/w	R/w			

Reset value = 00100000

Bit	Name	Function							
	tsrange[1:0]	Temperature Sensor Range Selection.							
		(FS range is 01024 mV)							
		00: -40° C 64° C (full operating range), with 0.5° C resolution (1 LSB in							
7: 6		the 8-bit ADC)							
		01: -40° C···· 85°C, with 1°C resolution (1 LSB in the 8-bit ADC)							
		11: 0 $^{\circ}$ C \cdots 85 $^{\circ}$ C, with 0.5 $^{\circ}$ C resolution (1 LSB in the 8-bit ADC)							
		10: -40° F \cdots 216° F, with 1 $^{\circ}$ F resolution (1 LSB in the 8-bit ADC)							
5	entsoffs	Temperature Sensor Offset to Convert from K to °C.							
4	entstrim	Temperature Sensor Trim Enable.							
3: 0	tstrim[3:0]	Temperature Sensor Trim Value.							

Register 13h. Temperature Value Offset

Bit	D7	D6	D5	D4	D3	D2	D1	D0			
Name		tvoffs[7:0]									
Туре		R/W									

Reset value = 00000000

Bit	Name	Function
7: 0	tuoffo[7:0]	Temperature Value Offset.
7:0	tvoffs[7:0]	This value is added to the measured temperature value. (MSB, tvoffs[8]: sign bit)

Note: If a new configuration is needed (e.g., for the WUT or the LDC), proper functionality is required. The function must first be disabled, then the settings changed, then enabled back on.



Register 14h. Wake-Up Timer Period 1

Bit	D7	D6	D5	D4	D3	D2	D1	D0	
Name	Reserved			wtr[4:0]					
Туре	R/w					R/w			

Reset value = xxx00011

Bit	Name	Function			
7: 5	Reserved	Reserved.			
		Wake Up Timer Exponent (R) Value*.			
4: 0	wtr[4:0]	Maximum value for R is decimal 20. A value greater than 20 will yield a result			
		as if 20 were written. R Value = 0 can be written here.			
*Note: The p	eriod of the wake-u	p timer can be calculated as Twut = $(4 \times M \times 2^R) / 32.768$ ms. R = 0 is allowed, and the			
maxir	maximum value for R is decimal 20. A value greater than 20 will result in the same as if 20 was written.				

Register 15h. Wake-Up Timer Period 2

Bit	D7	D6	D5	D4	D3	D2	D1	D0			
Name		wtm[15:8]									
Туре				R	/W						

Reset value = 00000000

Bit	Name	Function						
7: 0	wtm[15:8]	Wake Up Timer Mantissa (M) Value*.						
*Note: The p	*Note: The period of the wake-up timer can be calculated as $TwUT = (4 \times M \times 2^R) / 32.768$ ms.							

Register 16h. Wake-Up Timer Period 3

Bit	D7	D6	D5	D4	D3	D2	D1	D0				
Name		wtm[7:0]										
Туре		R/W										

Bit	Name	Function			
7: 0	wtm[7:0]	Wake Up Timer Mantissa (M) Value*.			
7:0	wtm[7:0]	M[7:0] = 0 is not valid here. Write at least decimal 1.			
*Note: The p	eriod of the wake-u	p timer can be calculated as Twut = (4 x M x 2^{R}) / 32.768 ms.			



Register 17h. Wake-Up Timer Value 1

Bit	D7	D6	D5	D4	D3	D2	D1	D0			
Name		wtm[15:8]									
Туре				R	2/W						

Reset value = xxxxxxxx

Bit	Name	Function						
7: 0	wtm[15:8]	Wake Up Timer Mantissa (M) Value*.						
*Note: The p	*Note: The period of the wake-up timer can be calculated as $TwuT = (4 \times M \times 2^R) / 32.768$ ms.							

Register 18h. Wake-Up Timer Value 2

Bit	D7	D6	D5	D4	D3	D2	D1	D0			
Name		wtm[7:0]									
Туре				R	/W						

Reset value = xxxxxxxx

Bit	Name	Function					
7: 0	wtm[7:0]	Wake Up Timer Mantissa (M) Value*.					
*Note: The p	*Note: The period of the wake-up timer can be calculated as Twut = $(4 \times M \times 2^R) / 32.768$ ms.						



Register 1Ah. Low Battery Detector Threshold

Bit	D7	D6	D5	D4	D3	D2	D1	D0	
Name	Reserved			lbdt[4:0]					
Туре		R				R/w			

Reset value = xxx10100

Bit	Name	Function			
7: 5	Reserved	Reserved.			
		Low Battery Detector Threshold.			
4: 0	lbdt[4:0]	This threshold is compared to Battery Voltage Level. If the Battery Voltage is			
		less than the threshold the Low Battery Interrupt is set. Default = 2.7 V.^*			
*Note: The th	*Note: The threshold can be calculated as Vthreshold = 1.7 + lbdt x 50 mV.				

Register 1Bh. Battery Voltage Level

Bit	D7	D6	D5	D4	D3	D2	D1	D0	
Name	Reserved			vbat[4:0]					
Туре	R					R			

Reset value = xxxxxxxx

Bit	Name	Function
7: 5	Reserved	Reserved.
		Battery Voltage Level.
4: 0	vbat[4:0]	The battery voltage is converted by a 5 bit ADC. In Sleep Mode the register is
		updated in every 1 s. In other states it measures continuously.



Register 30h. Data Access Control

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	Reserved	lsbfrst	crcdonly	Reserved	enpactx	encrc	crc[1:0]	
Туре	R/w	R/w	R/w	R/w	R/w	R/w	R/w	

Bit	Name	Function
7	Reserved	Reserved.
6	lsbfrst	LSB First Enable.
0	ISDIISL	The LSB of the data will be transmitted first if this bit is set.
5	orodonly	CRC Data Only Enable.
5	crcdonly	When this bit is set to 1 the CRC is calculated on the packet data fields only.
4	Reserved	Reserved.
		Enable Packet TX Handling.
	onnostv	If FIFO Mode (dtmod = 10) is being used automatic packet handling may be
3		enabled. Setting enpactx = 1 will enable automatic packet handling in the TX
5	enpactx	path. Register 30–4D allow for various configurations of the packet structure.
		Setting enpactx = 0 will not do any packet handling in the TX path. It will only
		transmit what is loaded to the FIFO.
2	oporo	CRC Enable.
2	encrc	Cyclic Redundancy Check generation is enabled if this bit is set.
		CRC Polynomial Selection.
		00: CCITT
1:0	crc[1:0]	01: CRC-16 (IBM)
		10: IEC-16
		11: Biacheva



Register 31h. EZMAC® Status

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name		Reserved						
Туре				R			R	R

Bit	Name	Function
7:2	Reserved	Reserved.
1	pktx	Packet Transmitting.
		When pktx = 1 the radio is currently transmitting a packet.
0	pksent	Packet Sent.
		A pksent = 1 a packet has been sent by the radio. (Same bit as in register 03,
		but reading it does not reset the IRQ)



Register 33h. Header Control 2

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	Reserved		hdlen[2:0]			syncle	en[1:0]	prealen[8]
Туре	R		R/w			R/	W	R/w

Bit	Name	Function				
7	Reserved	Reserved.				
		Header Length. Length of header used if packet handler is enabled for TX (enpactx). Headers				
6:4	hdlen[2:0]	are transmitted in descending order. 000: No TX header				
		001: Header 3 010: Header 3 and 2				
		011:Header 3 and 2 and 1100:Header 3 and 2 and 1 and 0				
		Fix Packet Length.				
3	fixpklen	When fixpklen = 1 the packet length (pklen[7:0]) is not included in the header. When fixpklen = 0 the packet length is included in the header.				
		Synchronization Word Length.				
		The value in this register corresponds to the number of bytes used in the				
		Synchronization Word. The synchronization word bytes are transmitted in				
2:1	ovrolon[1:0]	descending order.				
2.1	synclen[1:0]	00: Synchronization Word 3				
		01: Synchronization Word 3 and 2				
		10: Synchronization Word 3 and 2 and 1				
		11: Synchronization Word 3 and 2 and 1 and 0				
0	prealen[8]	MSB of Preamble Length.				
U	prealen[8]	See register Preamble Length.				



Register 34h. Preamble Length

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	prealen[7:0]							
Туре				F	R/W			

Reset value = 00001000

Bit	Name	Function
	prealen[7:0]	Preamble Length.
		The value in the prealen[8:0] register corresponds to the number of nibbles (4
		bits) in the packet. For example prealen[8:0] = '000001000' corresponds to a
7: 0		preamble length of 32 bits (8 x 4bits) or 4 bytes. The maximum preamble length
		is prealen[8:0] = 111111111 which corresponds to a 255 bytes Preamble.
		Writing 0 will have the same result as if writing 1, which corresponds to one
		single nibble of preamble.

Register 36h. Synchronization Word 3

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name		sync[31:24]						
Туре				R	/W			

Reset value = 00101101

Bit	Name	Function			
7 0	0.000[21:24]	Synchronization Word 3.			
7:0	7: 0 sync[31:24]	4 th byte of the synchronization word.			

Register 37h. Synchronization Word 2

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name		sync[23:16]						
Туре				R	/W			

Bit	Name	Function			
7.0	7: 0 sync[23:16]	Synchronization Word 2.			
7:0		3 rd byte of the synchronization word.			



Register 38h. Synchronization Word 1

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name		sync[15:8]						
Туре				F	R/W			

Reset value = 00000000

Bit	Name	Function			
7: 0	sync[15:8] Synchronization Word 1.				
7:0		2 nd byte of the synchronization word.			

Register 39h. Synchronization Word 0

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name		sync[7:0]						
Туре				R	/W			

Reset value = 00000000

Bit	Name	Function
7 0	7 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0	Synchronization Word 0.
7: 0	sync[7:0]	1 st byte of the synchronization word.

Register 3Ah. Transmit Header 3

Bit	D7	D6	D5	D4	D3	D2	D1	D0			
Name		txhd[31:24]									
Туре				R	/W						

Bit	Name	Function
7: 0	txhd[31:24]	Transmit Header 3.
7:0	tx11u[31.24]	4 th byte of the header to be transmitted.



Register 3Bh. Transmit Header 2

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name				txhd[2	23:16]			
Туре				F	R/W			

Reset value = 00000000

Bit	Name	Function			
7: 0	txhd[23:16]	Transmit Header 2.			
7:0		3 rd byte of the header to be transmitted.			

Register 3Ch. Transmit Header 1

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name				txhd[15:8]			
Туре				R	/W			

Reset value = 00000000

Bit	Name	Function
7 0	tub d[15,0]	Transmit Header 1.
7: 0	txhd[15:8]	2 nd byte of the header to be transmitted.

Register 3Ah. Transmit Header 3

Bit	D7	D6	D5	D4	D3	D2	D1	D0			
Name		txhd[7:0]									
Туре				R	/W						

Bit	Name	Function
7: 0	txhd[7:0]	Transmit Header 0.
7:0		1 st byte of the header to be transmitted.



Register 3Eh. Packet Length

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name				pkler	n[7:0]			
Туре				R	2/W			

Reset value = 00000000

Bit	Name	Function
		Packet Length.
		The value in the pklen[7:0] register corresponds directly to the number of bytes
7:0	pklen[7:0]	in the Packet. For example pklen[7:0] = '00001000' corresponds to a packet
7:0	pkieli[7.0]	length of 8 bytes. The maximum packet length is pklen[7:0] = '11111111', a 255
		byte packet. Writing 0 is possible, in this case we do not send any data in the
		packet.

Table 22. Internal Analog Signals Available on the Analog Test Bus

atb[3:0]	TESTp	TESTn
0	NC	NC
1	PLL_lcp_Test	PLL_IBG_5u
2	PLL_VBG_Bias	VSS_VCO
3	PLL_Vctrl	PLL_lptat_5u
4	VCO_LDO_VBG	VCO_LDO_VOUT
5	RF_LDO_VBG	RF_LDO_VOUT
6	PLL_LDO_VBG	PLL_LDO_VOUT
7	RCOSC_65kout	RCOSC_VSS
8	NC	NC
9	LBD_Comp	LBD_VBG(TS)
10	TS_VBG	TS_VTemp
11	DIG_LDO_VBG	DIG_LDO_VFB
12	PA_Ramp	NC
13	ADC_VIN	ADC_VDAC
14	NC	NC
15	NC	NC



Register 51h. Digital Test Bus Select

Bit	D7	D6	D5	D4	D3	D2	D1	D0	
Name	Reserved	ensctest	dtb[5:0]						
Туре	R/W	R/W			R	2/W			

Reset value = 00000000

Bit	Name	Function				
7	Reserved	Reserved.				
6	opostost	Scan Test Enable.				
0	ensctest	When set to 1 then GPIO0 will be the ScanEn input.				
5:0	dth[E:0]	Digital Test Bus.				
5:0	dtb[5:0]	GPIO must be configured to Digital Test Mux output.				

Table 23. Internal Digital Signals Available on the Digital Test Bus

dtb[4:0]	GPIO0	Signal	GPI01	Signal	GPIO2	Signal
0	wkup_clk_32k	wake-up 32kHz clock	rbase_en	first divided clock	clk_base	timebase clock
1	wkup_clk_32k	wake-up 32kHz clock	wake_up	wake-up event	tm1sec	1 sec timebase
2	ts_adc_en	aux. ADC enable	adc_rdy_n	aux. ADC conversion ready	adc_done	aux. ADC measurement done
3	cont_lbd	low battery continuous mode	lbd_on	low battery ON signal	bd	unfiltered output of LBD
4	div_clk_g	gated divided clock	uc_clk	microcontroller clock	ckout_rcsel	slow clock selected
5	en_div_sync	clock divider enable (sync'ed)	en_ckout	clock out enable	en_ckout_s	clock out enable (sync'ed)
6	osc30_en	oscillator enable	osc30_bias2x	oscillator bias control	xok	chip ready
7	xok	chip ready	zero_cap	cap. load zero	osc30_buff_en	buffer enable
8	tsadc_needed	aux. ADC enable	ext_retran	ext. retransmission request	tx_mod_gpio	TX modulation input
9	gpio_0_oen_n	GPIO0 output enable	gpio_0_aen	GPIO0 analog selection	gpio_0_aden	GPIO0 ADC input line enable
10	int_ack1	interrupt acknowledge 1	int_ack2	interrupt acknowledge 2	int_store	interrupt latch closed
11	ext_int2	ext. interrupt from GPIO2	irq_bit8	combined external status	msk_bit8	combined masked ext. int.
12	sdo_aux_sel	SDO aux. function select	sdo_aux	SDO aux. signal	nirq_aux_sel	nIRQ aux. function select
13	trdata_on_sdi	TX/RX data on SDI	tx_mod	TX modulation input	tx_clk_out	TX clock output
14	start_full_sync	RC osc. full calibration start	start_fine_sync	RC osc. fine calibration start	xtal_req	crystal req. for RC osc. cal.
15	coarse_rdy	RC osc. coarse cal. ready	fine_rdy	RC osc. fine cal. ready	xtal_req_sync	sync'ed crystal request
16	vco_cal_rst_s_n	VCO calibration reset	vco_cal	VCO calibration is running	vco_cal_done	VCO calibration done
17	vco_cal_en	VCO calibration enable	en_ref_cnt	reference counter enable	en_freq_cnt_s	frequency counter enable
18	vco_cal_en	VCO calibration enable	pos_diff	positive difference to goal	en_freq_cnt_s	frequency counter enable
19	dsm_clk_mux	DSM multiplexed clock	pll_fb_clk_tst	PLL feedback clock	pll_ref_clk_tst	PLL reference clock
20	dsm[0]	delta-sigma output	dsm[1]	delta-sigma output	dsm[2]	delta-sigma output
21	dsm[3]	delta-sigma output	pll_fbdiv15		dsm_rst_s_n	delta-sigma reset
22	pll_en	PLL enable: TUNE state	plit0_ok	PLL initial settling OK	plits_ok	PLL soft settling OK
23	ch_freq_req	frequency change request	plits_ok	PLL soft settling OK	vco_cal_done	VCO calibration done
24	vco_cal_en	VCO calibration enable	pll_vbias_shunt_en	VCO bias shunt enable	prog_req	frequency recalculation req.
25	bandgap_en	bandgap enable	frac_div_en	fractional divider enable	buff3_en	buffer3 enable
26	pll_pfd_up	PFD up signal	pll_pfd_down	PFD down signal	pfd_up_down	PFD output change (XOR'ed
27	pll_lock_detect	PLL lock detect	p <mark>ll_</mark> en	PLL enable: TUNE state	plit0_ok	PLL initial settling OK
28	pll_en	PLL enable: TUNE state	pll_lock_detect	PLL lock detect	plits_ok	PLL soft settling OK
29	pwst[0]	internal power state	pwst[1]	internal power state	pwst[2]	internal power state



Table 23. Internal Digital Signals Available on the Digital Test Bus (Continued)

dtb[4:0]	GPIO0	Signal	GPI01	Signal	GPIO2	Signal
30	xok	chip ready: READY state	pll_en	PLL enable: TUNE state	tx_en	TX enable: TX state
31	ts_en	temperature sensor enable	auto_tx_on	automatic TX ON	tx_off	TX OFF
32	ch_freq_req	frequency change request	retum_tx	retum from TX	pk_sent	packet sent
33	retran_req	retransmission request	tx_ffpt_store	TX FIFO pointer store	tx_ffpt_restore	TX FIFO pointer restore
34	pa_on_trig	PA ON trigger	dly_5us_ok	5 us delay expired	mod_dly_ok	modulator delay expired
35	tx_shdwn	TX shutdown	ramp_start	modulator ramp down start	ramp_done	modulator ramp down ended
36	pk_sent_dly	delayed packet sent	tx_shdwn_done	TX shutdown done	pa_ramp_en	PA ramp enable
37	tx_en	TX enable: TX state	Ido_rf_precharge	RF LDO precharge	pa_ramp_en	PA ramp enable
38	pa_on_trig	TX enable: TX state	dp_tx_en	packet handler (TX) enable	mod_en	modulator enable
39	reg_wr_en	register write enable	reg_rd_en	register rdead enable	addr_inc	register address increment
40	dp_tx_en	packet handler (TX) enable	data_start	start of TX data	pk_sent	packet has been sent
41	data_start	start of TX data	tx_out	packet handler TX data out	pk_sent	packet has been sent
42	ramp_done	ramp is done	data_start	start of TX data	pk_tx	packet is being transmitted
43	tx_ffaf	TX FIFO almost full	tx_fifo_wr_en	TX FIFO write enable	tx_ffem_tst	internal TX FIFO empty
44	clk_mod	modulator gated 10MHz clock	tx_clk	TX clock from NCO	rd_clk_x8	read clock = tx_clk / 10
45	mod_en	modulator enable	ramp_start	start modulator ramping down	ramp_done	modulator ramp done
46	data_start	data input start from PH	ook_en	OOK modulation enable	ook (also internal PN9)	OOK modulation
47	prog_req	freq. channel update request	freq_err	wrong freq. indication	dsm_rst_s_n	dsm sync. reset
48	mod_en	modulator enable	tx_rdy	TX ready	tx_clk	TX clock from NCO
49	dp_rx_en	packet handler (RX) enable	prea_valid	valid preamble	pk_srch	packet is being searched
50	pk_srch	packet is being searched	sync_ok	sync. word has been detected	rx_data	packet handler RX data input
51	pk_rx	packet is being received	sync_ok	sync. word has been detected	pk_valid	valid packet received
52	sync_ok	sync. word has been detected	crc_error	CRC error has been detected	hdch_error	header error detected
53	direct_mode	direct mode	rx_ffaf	RX FIFO almost full	rx_fifo_rd_en	RX FIFO read enable
54	bit_clk	bit clock	prea_valid	valid preamble	rx_data	demodulator RX data output
55	prea_valid	valid preamble	prea_inval	invalid preamble	ant_div_sw	antenna switch (algorythm)
56	sync_ok	sync. word has been detected	bit_clk	bit clock	rx_data	demodulator RX data output
57	demod phase[4]	demodulator phase MSB	demod phase [3]	demodulator MSB-1	demod phase [2]	demodulator MSB-2
58	prea_valid	valid preamble	demod_tst[2]	demodulator test	demod_tst[1]	demodulator test
59	agc_smp_clk	AGC sample clock	win_h_tp	window comparator high	win_l_tp	window comparator low dly'd
60	agc_smp_clk	AGC sample clock	win_h_dly_tp	window comparator high	win_l_dly_tp	window comparator low dly'd
61	ldc_on	active low duty cycle	pll_en	PLL enable: TUNE state	rx_en	RX enable: RX state
62	ldc_on	active low duty cycle	no_sync_det	no sync word detected	prea_valid	valid preamble
63	adc_en	ADC enable	adc_refdac_en	ADC reference DAC enable	adc_rst_n	combined ADC reset



Register 52h. TX Ramp Control

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	Reserved		txmod[2:0]		ldoramp[1:0]		txramp[1:0]	
Туре	R/w		R/w		R/w		R/w	

Bit	Name	Function
7	Reserved	Reserved.
		TX Modulation Delay.
6:4	txmod[2:0]	The time delay between PA enable and the beginning of the TX modulation to
0.4	txinou[2.0]	allow for PA ramp-up. It can be set from 0 μs to 28 μs in 4 μs steps. This also
		works during PA ramp down.
		TX LDO Ramp Time.
		The RF LDO is used to help ramp the PA to prevent VCO pulling and spectral
		splatter.
3:2	ldoramp[1:0]	00: 5 μs
		01: 10 μs
		10: 15 µs
		11: 20 μs
		TX Ramp Time.
		The PA is ramped up slowly to prevent VCO pulling and spectral splatter. This
		register sets the time the PA is ramped up.
1:0	txramp[1:0]	00: 5 μs
		01: 10 µs
		10: 15 μs
		11: 20 μs



The total settling time (cold start) of the PLL after the calibration can be calculated as Tcs = Ts + To.

Register 53h. PLL Tune Time

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name			pllts[4:0]	pllts0				
Туре			R/w		R/w			

Reset value = 01010010

Bit	Name	Function
		PLL Soft Settling Time (Ts).
		This register will set the settling time for the PLL from a previous locked
7:3	pllts[4:0]	frequency in Tune mode. The value is configurable between 0 μs and 310 $\mu s,$
		in 10 μ s intervals. The default plltime corresponds to 100 μ s. See formula
		above.
		PLL Settling Time (To).
2:0	allta Ot	This register will set the time allowed for PLL settling after the calibrations are
2.0	pllts0t	completed. The value is configurable between 0 μ s and 70 μ s, in 10 μ s steps.
		The default pllt0 corresponds to 20 µs. See formula above.

Register 54h. PA Boost

Bit	D7	D6	D5 D4 D3 D2				D1	D0
Name	Reserv	/ed[7:6]	inv_pre_th				ldo_pa_boost	pa_vbias_boost
Туре	F	R/W	R/w			R/w	R/w	

Reset value = 01010100

Bit	Name	Function
7:6	Reserved[7:6]	Reserved.
5:2	inv_pre_th[5:2]	Invalid Preamble Threshold.
1	ldo_pa_boost	LDO PA Boost.
0	pa_vbias_boost	PA VBIAS Boost.

Invalid preamble will be evaluated during this period: (invalid_preamble_Threshold x 4) x Bit Rate period.



Register 55h. Calibration Control

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	Reserved	xtalstarthalf	Reserved	enrcfcal	rccal	vcocaldp	vcocal	skipvco
Туре	R	R/w	R	R/w	R/w	R/w	R/w	R/w

Reset value = x1x00100

Bit	Name	Function
7	Reserved	Reserved.
6	xtalstarthalf	If Set, the Xtal Wake Time Period is Halved.
5	Reserved	Reserved.
4	enrcfcal	RC Oscillator Fine Calibration Enable.
4	enicical	If this bit is set to 1 then the RC oscillator performs fine calibration in every app. 30 s.
		RC Calibration Force.
		If setting rccal = 1 will automatically perform a forced calibration of the 32 kHz RC
3	rccal	Oscillator. The RC OSC will automatically be calibrated if the Wake-Up-Timer is
5	rccar	enabled. The calibration takes 2 ms. The 32 kHz RC oscillator must be enabled to
		perform a calibration. Setting this signal from a 0 to 1 will initiate the calibration. This bit
		is cleared automatically.
		VCO Calibration Double Precision Enable.
2	vcocaldp	When this bit is set to 1 then the VCO calibration measures longer thus
		calibrates more precisely.
		VCO Calibration Force.
1	vcocal	If in Idle Mode and pllon = 1, setting vcocal = 1 will force a one time calibration
		of the synthesizer VCO. This bit is cleared automatically.
		Skip VCO Calibration.
0	skipvco	Setting skipvco = 1 will skip the VCO calibration when going from the Idle state
		to the TX state.



Register 56h. Modem Test

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	bcrfbyp	slicfbyp	dttype	oscdeten	ookth	refclksel	refclkinv	distogg
Туре	R/w	R/w	R/w	R/w	R/w	R/w	R/w	R/w

Reset value = 00000000

Bit	Name	Function
7	bcrfbyp	If set, BCR phase compensation will be bypassed.
6	slicfbyp	If set, slicer phase compensation will be bypassed.
		Dithering Type.
5	dttype	If low and dither enabled, we add +1/0, otherwise if high and dithering enabled,
		we add ±1.
4	oscdeten	If low, the ADC Oscillation Detection mechanism is allowed to work. If set, we
4	USCUEIEIT	disable the function.
3	ookth	If set, in OOK mode, the slicer threshold will be estimated by 8 bits of preamble.
3	OOKIT	By default, this bit is low and the demod estimate the threshold after 4 bits.
		Delta-Sigma Reference Clock Source Selection
2	refclksel	1: 10 MHz
		0: PLL
1	refclkinv	Delta-Sigma Reference Clock Inversion Enable.
0	distogg	If reset, the discriminator toggling is disabled.

Register 57h. Charge Pump Test

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	pfdrst	fbdiv_rst	cpforceup	cpforcedn	cdonly	cdcurr[2:0]		
Туре	R/w	R/w	R/w	R/w	R/w	R/w		

Bit	Name	Function
7	pfdrst	Direct Control to Analog.
6	fbdiv_rst	Direct Control to Analog.
5	cpforceup	Charge Pump Force Up.
4	cpforcedn	Charge Pump Force Down.
3	cdonly	Charge Pump DC Offset Only.
2:0	cdcurr[2:0]	Charge Pump DC Current Selection.



Register 58h. Charge Pump Current Trimming/Override

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	срс	urr[1:0]	cpcorrov		cporr[4:0]			
Туре	F	R/W	R/w			R/w		

Reset value = 100xxxxx

Bit	Name	Function
		Charge Pump Current (Gain Setting).
7:6	cpcurr[1:0]	Changing these bits will change the BW of the PLL. The default setting is
		adequate for all data rates.
5	cpcorrov	Charge Pump Correction Override Enable.
		Charge Pump Correction Value.
4:0	oporr[4:0]	During read, you read what the Charge Pump sees. If cpcorrov = 1, then the
4.0	cporr[4:0]	value you write will go to the Charge Pump, and will also be the value you read.
		By default, cpcorr[4:0] wakes up as all Zeros.

Register 59h. Divider Current Trimming/Delta-Sigma Test

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	txcorboosten	fbdivhc	d3trim[1:0]		d2trim[1:0]		d1p5tri	im[1:0]
Туре	R/w	R/w	R/w		R/w		R/v	N

Bit	Name	Function
7	txcorboosten	If this is Set, then vcocorr (reg 5A[5:2]) = 1111 during TX Mode and VCO
/	IXCOIDOOSIEII	CAL followed by TX.
6	fbdivhc	Feedback (fractional) Divider High Current Enable (+5 μA).
5:4	d3trim[1:0]	Divider 3 Current Trim Value.
3:2	d2trim[1:0]	Divider 2 Current Trim Value.
1:0	d1p5trim[1:0]	Divider 1.5 (div-by-1.5) Current Trim Value.



Register 5Ah. VCO Current Trimming

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	txcurboosten	vcocorrov	vcocorr[3:0]				vcocu	ır[1:0]
Туре	R/w	R/w		R/w R/w			N	

Reset value = 00000011

Bit	Name	Function
7.		If this is Set, then vcocur = 11 during TX Mode and VCO CAL followed by
7.	txcurboosten	TX.
6	vcocorrov	VCO Current Correction Override.
5:2	vcocorr[3:0]	VCO Current Correction Value.
1:0	vcocur[1:0]	VCO Current Trim Value.

Register 5Bh. VCO Calibration/Override

Bit	D7	D6	D5	D4	D3	D2	D1	D0	
Name	vcocalov/vcdone	vcocal[6:0]							
Туре	R/w				R/w				

Bit	Name	Function
		VCO Calibration Override/Done.
		When vcocalov = 0 the internal VCO calibration results may be viewed by
7.	vcocalov/vcdone	reading the vcocal register. When vcocalov = 1 the VCO results may be
		overridden externally through the SPI by writing to the vcocal register. Reading
		this bit gives 1 if the calibration process has been finished.
6:0	vcocal[6:0]	VCO Calibration Results.



Register 5Ch. Synthesizer Test

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	dsmdt	vcotype	enoloop	dsmod	dsorder[1:0]		dsrstmode	dsrst
Туре	R/w	R	R/w	R/w	R/w		R/w	R/w

Reset value = 0x001110

Bit	Name	Function
7	dsmdt	Enable DSM Dithering.
1	usmut	If low, dithering is disabled.
		VCO Type.
6	vcotype	0: basic, constant K
		1: single varactor, changing K
5	enoloop	Open Loop Mode Enable.
		Delta-Sigma Modulus.
4	dsmod	0: 64 000
		1: 65 536
		Delta-Sigma Order.
		00: 0 order
3:2	dsorder[1:0]	01: 1 st order
		10: 2 nd order
		11: Mash 111
1	dsrstmode	Delta-Sigma Reset Mode.
0	dsrst	Delta-Sigma Reset.



Register 5Dh. Block Enable Override 1

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	enmix	enina	enpga	enpa	enbf5	endv32	enbf12	enmx2
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Reset value = 00000000

Bit	Name	Function
7	enmix	Mixer Enable Override.
6	enina	LNA Enable Override.
5	enpga	PGA Enable Override.
4	enpa	Power Amplifier Enable Override.
3	enbf5	Buffer 5 Enable Override.
2	endv32	Divider 3_2 Enable Override.
1	enbf12	Buffer 1_2 Enable Override.
0	enmx2	Multiplexer 2 Enable Override.

Register 5Eh. Block Enable Override 2

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	ends	enldet	enmx3	enbf4	enbf3	enbf11	enbf2	pllreset
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Bit	Name	Function
7	ends	Delta-Sigma Enable Override.
6	opldot	Lock Detect Enable.
0	6 enldet	(direct control, does not need override!)
5	enmx3	Multiplexer 3 Enable Override.
4	enbf4	Buffer 4 Enable Override.
3	enbf3	Buffer 3 Enable Override.
2	enbf11	Buffer 1_1 Enable Override.
1	enbf2	Buffer 2 Enable Override.
0	pllreset	PLL Reset Enable Override.



Register 5Fh. Block Enable Override 3

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	enfrdv	endv31	endv2	endv1p5	dvbshunt	envco	encp	enbg
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Bit	Name	Function
7	enfrdv	Fractional Divider Enable Override.
6	endv31	Divider 3_1 Enable Override.
5	endv2	Divider 2 Enable Override.
4	endv1p5	Divider 1.5 (div-by-1.5) Enable Override.
3	dvbshunt	VCO Bias Shunt Enable Override Mode.
2	envco	VCO Enable Override.
1	encp	Charge Pump Enable Override.
0	enbg	Bandgap Enable Override.



Register 62h. Crystal Oscillator/Power-on-Reset Control

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	pwst[2:0]			clkhyst	enbias2x	enamp2x	bufovr	enbuf
Туре	R			R/W	R/W	R/W	R/W	R/W

Reset value = xxx00100

Bit	Name	Function
		Internal Power States of the Module.
		LP: 000
7:5	pwst[2:0]	RDY: 001
		Tune: 011
		TX: 010
4	clkhyst	Clock Hysteresis Setting.
3	enbias2x	2 Times Higher Bias Current Enable.
2	enamp2x	2 Times Higher Amplification Enable.
		Output Buffer Enable Override.
1	bufovr	If set to 1 then the enbuf bit controls the output buffer.
I	DUIOVI	0: output buffer is controlled by the state machine.
		1: output buffer is controlled by the enbuf bit.
0	onbuf	Output Buffer Enable.
U	enbuf	This bit is active only if the bufovr bit is set to 1.



Register 63h. RC Oscillator Coarse Calibration/Override

Bit	D7	D6	D5	D4	D3	D2	D1	D0		
Name	rccov		rcc[6:0]							
Туре	R/W				R/W					

Reset value = 00000000

Bit	Name	Function					
	7 rccov	RC Oscillator Coarse Calibration Override.					
7		When rccov = 0 the internal Coarse Calibration results may be viewed by					
1		reading the rcccal register. When rccov = 1 the Coarse results may be					
		overridden externally through the SPI by writing to the rcccal register.					
6:0	rcc[6:0]	RC Oscillator Coarse Calibration Override Value/Results.					

Register 64h. RC Oscillator Fine Calibration/Override

Bit	D7	D6	D5	D4	D3	D2	D1	D0		
Name	rcfov		rcf[6:0]							
Туре	R/W				R/W					

Bit	Name	Function				
	7 rcfov	RC Oscillator Fine Calibration Override.				
7		When rcfov = 0 the internal Fine Calibration results may be viewed by reading				
1		the rcfcal register. When rcfov = 1 the Fine results may be overridden externally				
		through the SPI by writing to the rcfcal register.				
6:0	rcf[6:0]	RC Oscillator Fine Calibration Override Value/Results.				



Register 65h. LDO Control Override

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	enspor	enbias	envcoldo	enifldo	enrfldo	enpllldo	endigldo	endigpwdn
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Reset value = 10000001

Bit	Name	Function
7	enspor	Smart POR Enable.
6	enbias	Bias Enable.
5	envcoldo	VCO LDO Enable.
4	enifldo	IF LDO Enable.
3	enrfldo	RF LDO Enable.
2	enpIIIdo	PLL LDO Enable.
1	endigldo	Digital LDO Enable.
0	endigpwdn	Digital Power Domain Powerdown Enable in Idle Mode.

Register 66h. LDO Level Settings

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	enovr	enxtal	ents	enrc32	Reserved	diglvl		
Туре	R/W	R/W	R/W	R/W	R		R/W	

Bit	Name	Function
		Enable Overrides.
7	enovr	If high, ovr values are output to the blocks and can enable or disable them, if
		low, some ovr value can only enable the blocks.
6	enxtal	Xtal Override Enable Value.
5	ents	Temperature Sensor Enable.
4	enrc32	32K Oscillator Enable.
3	Reserved	Reserved.
2:0	diglvl	Digital LDO Level Setting.



Register 6Bh. GFSK FIR Filter Coefficient Address

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name			Reserved	firadd[2:0]				
Туре			R				R/W	

Reset value = xxxxx000

Bit	Name	Function
7:3	Reserved	Reserved.
		GFSK FIR Filter Coefficient Look-up Table Address.
		The address for Gaussian filter coefficients used in the TX path. The default
		GFSK setting is for BT = 0.5. It is not needed to change or load the GFSK
		Coefficients if BT = 0.5 is satisfactory for the system.
		000: i_coe0 (Default = d1)
2:0	firadd[2:0]	001: i_coe1 (Default = d3)
		010: i_coe2 (Default = d6)
		011: i_coe3 (Default = d10)
		100: i_coe4 (Default = d15)
		101: i_coe5 (Default = d19)
		110: i_coe6 (Default = d20)

Register 6Ch. GFSK FIR Filter Coefficient Value

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	Res	erved	firval[5:0]					
Туре	R/W				R	/W		

Reset value = xxxxx000

Bit	Name	Function
7:6	Reserved	Reserved.
5:0	firval[5:0]	FIR Coefficient Value in the IOok-up Table Addressed by the firadd[2:0].
5.0	5:0 IIIvai[5:0]	The default coefficient can be read or modified.



Register 6Dh. TX Power

Bit	D7	D6	D5	D4	D3	D2	D1	D0	
Name		Res	erved		lna_sw	txpow[2:0]			
Туре		R			R/W		R/W		

Reset value = xxxx1000

Bit	Name	Function
7:4	Reserved	Reserved.
		LNA Switch Controller.
3	lna_sw	If set, Ina_sw control from the digital will go high during TX modes, and low
		during other times. If reset, the digital control signal is low at all times.
2:0	typoy/2:01	TX Output Power.
2:0	txpow[2:0]	The output power is configurable in ~3 dBm steps.



Register 6Eh. TX Data Rate 1

Bit	D7	D6	D5	D4	D3	D2	D1	D0			
Name		txdr[15:8]									
Туре				R	z/W						

Reset value = 00001010

Bit	Name	Function			
7:0		Data Rate Upper Byte.			
7.0	txdr[15:8]	See formula above.			

The data rate can be calculated as: $TX_DR = 10^3 \times txdr[15:0] / 2^{16}$ [kbps] (if address 70[5] = 0) <u>or</u> The data rate can be calculated as: $TX_DR = 10^3 \times txdr[15:0] / 2^{21}$ [kbps] (if address 70[5] = 1)

Register 6Fh. TX Data Rate 0

Bit	D7	D6	D5	D4	D3	D2	D1	D0			
Name		txdr[7:0]									
Туре				R	/W						

Bit	Name	Function
7:0	tvdr[7:0]	Data Rate Lower Byte.
7.0	7:0 txdr[7:0]	See formula above. Defaults = 40 kbps.



Register 70h. Modulation Mode Control 1

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	Reserved		txdtrtscale	enphpwdn	manppol	enmaninv	enmanch	enwhite
Туре	R		R/W	R/W	R/W	R/W	R/W	R/W

Bit	Name	Function
7:6	Reserved	Reserved.
5	txdtrtscale	This bit should be set for Data Rates below 30 kbps.
4	enphpwdn	If set, the Packet Handler will be powered down when module is in low power
4	enpripwuri	mode.
		Manchester Preamble Polarity (will transmit a series of 1 if set, or series of 0 if
3	mannal	reset).
3	manppol	This bit affects ONLY the transmitter side, not the receiver. This is valid ONLY if
		Manchester Mode is enabled.
2	enmaninv	Manchester Data Inversion is Enabled if this bit is set.
1	enmanch	Manchester Coding is Enabled if this bit is set.
0	enwhite	Data Whitening is Enabled if this bit is set.



Register 71h. Modulation Mode Control 2

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	trclk	trclk[1:0]		dtmod[1:0]		fd[8]	modty	/p[1:0]
Туре	R/\	R/W		N	R/W	R/W	R	/W

Reset value = 00000000

Bit	Name	Function					
		TX Data Clock Configuration.					
		00: No TX Data CLK is available (asynchronous mode – Can only work with					
		modulations FSK or OOK).					
7:6	trclk[1:0]	01: TX Data CLK is available via the GPIO (one of the GPIO's should be					
		programmed as well).					
		10: TX Data CLK is available via the SDO pin.					
		11: TX Data CLK is available via the nIRQ pin.					
		Modulation Source.					
	dtmod[1:0]	00: Direct Mode using TX_Data function via the GPIO pin (one of the GPIO's					
5:4		should be programmed accordingly as well)					
5.4		01: Direct Mode using TX_Data function via the SDI pin (only when nSEL is high)					
		10: FIFO Mode					
		11: PN9 (internally generated)					
3	eninv	TX Data.					
2	fd[9]	MSB of Frequency Deviation Setting, see "Register 72h. Frequency					
2	fd[8]	Deviation".					
		Modulation Type.					
		00: Unmodulated carrier					
1: 0	modtyp[1:0]	01: OOK					
		10: FSK					
		11: GFSK (enable TX Data CLK (trclk[1:0]) when direct mode is used)					

The frequency deviation can be calculated: Fd = 625 Hz x fd[8:0].



Register 72h. Frequency Deviation

Bit	D7	D6	D5	D4	D3	D2	D1	D0			
Name		fd[7:0]									
Туре				R	/W						

Reset value = 00100000

Bit	Name	Function
7:0	fd[7:0]	Frequency Deviation Setting.
7.0	10[7.0]	See formula above.

Note: It's recommended to use modulation index of 1 or higher (maximum allowable modulation index is 32). The modulation index is defined by 2FN/FR were FD is the deviation and RB is the data rate. When Manchester coding is enabled the modulation index is defined by FD/RB.

Register 73h. Frequency Offset 1

Bit	D7	D6	D5	D4	D3	D2	D1	D0			
Name		fo[7:0]									
Туре				R	W						

Bit	Name	Function
		Frequency Offset Setting.
7:0	fo[7:0]	The frequency offset can be calculated as Offset = 156.25 Hz x (hbsel + 1) x fo[7:0].
		fo[9:0] is a twos complement value.



Register 74h. Frequency Offset 2

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name			fo[9	9:8]				
Туре				R			R/	W

Reset value = 00000000

Bit	Name	Function
7:2	Reserved	Reserved.
		Upper Bits of the Frequency Offset Setting.
1:0	fo[9:8]	fo[9] is the sign bit. The frequency offset can be calculated as Offset = 156.25 Hz x
		(hbsel + 1) x fo[7:0]. fo[9:0] is a twos complement value.

Register 75h. Frequency Band Select

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	Reserved	sbsel	hbsel	fb[4:0]				
Туре	R	R/W	R/W			R/W		

Reset value = 01110101

Bit	Name	Function
7	Reserved	Reserved.
6	sbse	Side Band Select.
		High Band Select.
5	hbsel	Setting hbsel = 1 will choose the frequency range from 480–930 MHz (high bands). Setting
		hbsel = 0 will choose the frequency range from 240–479.9 MHz (low bands).
		Frequency Band Select.
		Every increment corresponds to a 10 MHz Band for the Low Bands and a 20 MHz Band
4:0	fb[4:0]	for the High Bands. Setting fb[4:0] = 00000 corresponds to the 240–250 MHz Band for
		hbsel = 0 and the 480–500 MHz Band for hbsel = 1. Setting fb[4:0] = 00001 corresponds
		to the 250–260 MHz Band for hbsel = 0 and the 500–520 MHz Band for hbsel = 1.

The RF carrier frequency can be calculated as follows:

 $f_{carrier} = (f_b+24+(f_c+f_o) / 64000) \times 10000 \times (hbsel+1) + (f_{hch} \times f_{hs} \times 10) [kHz],$

where parameters f_c , f_o , f_b and hb_sel come from registers 73h–77h. Parameters f_{hch} and f_{hs} come from register 79h and 7Ah.



Register 76h. Nominal Carrier Frequency

Bit	D7	D6	D5	D4	D3	D2	D1	D0		
Name		fc[15:8]								
Туре				R	W					

Reset value = 10111011

Bit	Name	Function
7:0	7.0 6-[45:0]	Nominal Carrier Frequency Setting.
7.0	fc[15:8]	See formula above.

Register 77h. Nominal Carrier Frequency

Bit	D7	D6	D5	D4	D3	D2	D1	D0			
Name		fc[7:0]									
Туре				R	W						

Reset value = 10000000

Bit	Name	Function
7:0	fo[7:0]	Nominal Carrier Frequency Setting.
7:0	fc[7:0]	See formula above.

Register 78h. Miscellaneous Settings

Bit	D7	D6	D5	D4	D3	D2	D1	D0		
Name		Reserve	ed[7:4]		Alt_PA_Seq		rcosc_cal[2:0]			
Туре		R/W					R/W			

Bit	Name	Function
7:0	Reserved[7:4]	Reserved.
3		Alternative PA sequencing.
3	Alt_PA_Seq	If set, we will enable the alternative PA sequence. By default, this is not enabled.
		rcosc_cal[2:0].
2:0	rcosc_cal[2:0]	Fine changes on the RC OSC Calibration target frequency, to help compensate for
		"calibration biases." This register should not be changed by costumers.



Register 79h. Frequency Hopping Channel Select

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name				fhch	[7:0]			
Туре				R	/W			

Reset value = 00000000

Bit	Name	Function
7:0	fhch[7:0]	Frequency Hopping Channel Number.

Register 7Ah. Frequency Hopping Step Size

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name				fhs	7:0]			
Туре				R	W			

Bit	Name	Function			
		Frequency Hopping Step Size in 10 kHz Increments.			
7:0	fhs[7:0]	See formula for the nominal carrier frequency at "Register 76h. Nominal Carrier			
		Frequency".			



Register 7Bh. Turn Around and 15.4 Length Compliance

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	15.4 Length		Reserved[6:3]			turn_around_en	phase[1:0]	
Туре	R/W		R/W		R/W	R/W		

Reset value = 01111011

Bit	Name	Function
7	15.4 Length	15.4 Packet Length Compliance. If set, then PK Length definition for both TX and RX will also include the CRC bytes, If reset, then the Length refers ONLY to the DATA payload. For example, writing "9" to this register when it is set, means we are sending/expecting "7" bytes of DATA,
		and the other "2" should be the CRC (CRC should be enabled separately).
6:3	Reserved[6:3]	Reserved.
2	turn_around_en	Turn Around Enable. Enabling for the turn around functionality.
1:0	phase[1:0]	Turn Around Phase. The RX to TX and vice-versa change in frequency will happen (if bit [2] is set) at the last byte, and these two registers set the bit position in which the frequency shifts should occur. Make sure it does not happen to early otherwise the last bits will be missed.

Register 7Ch. TX FIFO Control 1

Bit	D7	D6	D5	D4	D3	D2	D1	D0	
Name	Reserved		txafthr[5:0]						
Туре	R/W		R/W						

Bit	Name	Function
7: 6	Reserved	Reserved.
5: 0	txafthr[5:0]	TX FIFO Almost Full Threshold.



Register 7Dh. TX FIFO Control 2

Bit	D7	D6	D5	D4	D3	D2	D1	D0
Name	Rese	erved	txfaethr[5:0]					
Туре	R/V	V	R/W					

Reset value = 00000100

Bit	Name	Function
7: 6	Reserved	Reserved.
5: 0	txfaethr[5:0]	TX FIFO Almost Empty Threshold.

Register 7Fh. FIFO Access

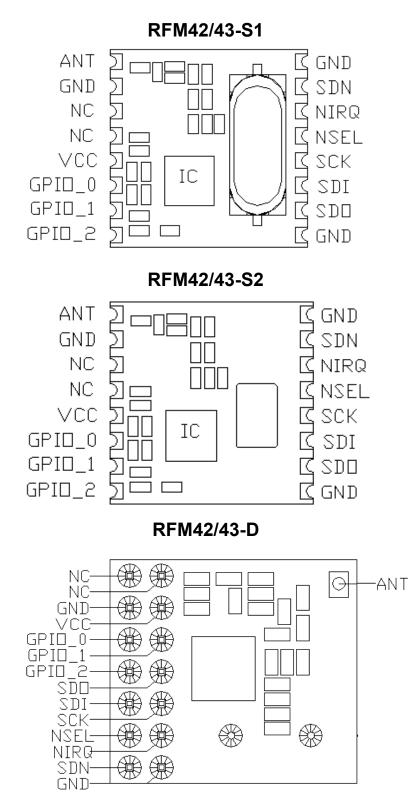
Bit	D7	D6	D5	D4	D3	D2	D1	D0	
Name		fifod[7:0]							
Туре		R/W							

Reset value = NA

Bit	Name	Function
		FIFO Data.
		A Write (R/W = 1) to this Address will begin a Burst Write to the TX FIFO. The
7: 0	fifod[7:0]	FIFO will be loaded in the same manner as a Burst SPI Write but the SPI address
		will not be incremented. To conclude the TX FIFO Write the SEL pin should be
		brought HIGH, in the same manner.



11. Pin Descriptions: RFM42/43





VCC	S	+1.8 to +3.6 V supply voltage. The recommended Vcc supply voltage is +3.3 V.
GND	S	Ground reference.
GPIO_0	I/O	General Purpose Digital I/O that may be configured through the registers to perform various
GPIO_1	I/O	functions including: Microcontroller Clock Output, FIFO status, POR, Wake-Up timer, Low
		Battery Detect, TRSW, AntDiversity control, etc. See the SPI GPIO Configuration Registers,
GPIO_2	I/O	Address 0Bh, 0Ch, and 0Dh for more information.
600	0	0-Vcc V digital output that provides a serial readback function of the internal control
SDO	0	registers.
SDI		Serial Data input. 0-Vcc V digital input. This pin provides the serial data stream for the 4-line
501	1	serial data bus.
00116		Serial Clock input. 0-VDD V digital input. This pin provides the serial data clock function for
SCLK	1	the 4-line serial data bus. Data is clocked into the RFM42/43 on positive edge transitions.
-051		Serial Interface Select input. 0- Vcc V digital input. This pin provides the Select/Enable
nSEL	1	function for the 4-line serial data bus. The signal is also used to signify burst read/write mode.
		General Microcontroller Interrupt Status output. When the RFM42/43 exhibits anyone of the
		Interrupt Events the nIRQ pin will be set low=0. Please see the Control Logic registers
nIRQ	0	section for more information on the Interrupt Events. The Microcontroller can then determine
		the state of the interrupt by reading a corresponding SPI Interrupt Status Registers, Address
		03h and 04h.
	I	Shutdown input pin. 0–Vcc V digital input. SDN should be = 0 in all modes except Shutdown
SDN		mode. When SDN =1 the chip will be completely shutdown and the contents of the registers
		will be lost.
ANT	I/O	RF signal output/input.(50 OHM output /input Impedance)
NC		No Connection

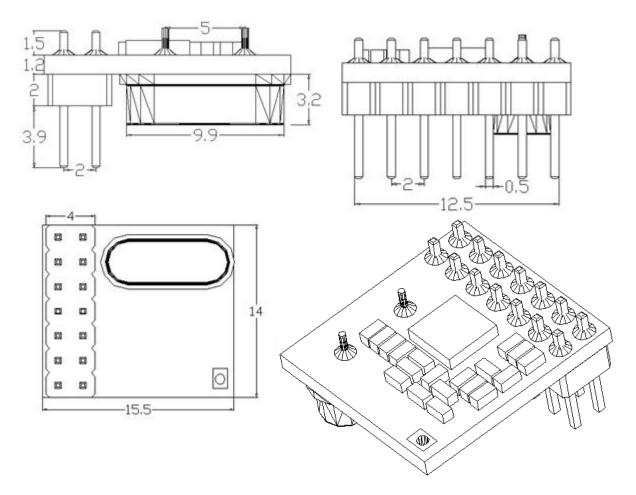


12. Mechanical Dimension:RFM42/43

SMD PACKAGE (S1) -16 1.3 3,9 16 ø1. 13 1.0 IC 3.9 60 A REAL PROPERTY. INVE INVE INVE INVE WM WAL WA m SMD PACKAGE (S2) 16



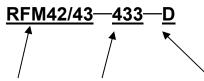
DIP PACKAGE (D)





13. Ordering Information

Part Number=module type—operation band—package type



module type operation band Package

example: 1, RFM43 module at 433MHz band, DIP : RFM43-433-D。

2, RFM42 module at 868MHZ band, SMD, thickness at 4.9mm: RFM42-868-S1.



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