## Features

- Micro power design
- 2.4 V to 5.5 V battery operation
- High sensitivity and high stability of the magnetic switching points
- High resistance to mechanical stress
- Digital output signal
- Switching for both poles of a magnet (omnipolar)
- Programming pin for the switching direction of the output

- Not suitable for automotive application


## Functional Description

The TLE 4917 is an Integrated Hall-Effect Sensor designed specifically to meet the requirements of low-power devices. e.g. as an On/Off switch in Cellular Flip-Phones, with battery operating voltages of $2.4 \mathrm{~V}-5.5 \mathrm{~V}$.

Precise magnetic switching points and high temperature stability are achieved through the unique design of the internal circuit.

An onboard clock scheme is used to reduce the average operating current of the IC.
During the operate phase the IC compares the actual magnetic field detected with the internally compensated switching points. The output $Q$ is switched at the end of each operating phase.

During the Stand-by phase the output stage is latched and the current consumption of the device reduced to some $\mu \mathrm{A}$.

The IC switching behaviour is omnipolar, i.e. it can be switched on with either the North or South pole of a magnet.

The PRG pin can be connected to $\mathrm{V}_{\mathrm{S}}$ which holds the output $\mathrm{V}_{\mathrm{Q}}$ at a High level for $\mathrm{B}=0 \mathrm{mT}$; conversely the output $\mathrm{V}_{\mathrm{Q}}$ can be inverted by connecting the PRG pin to GND, which will hold the output $\mathrm{V}_{\mathrm{Q}}$ at a Low level for $\mathrm{B}=0 \mathrm{mT}$. In this later case the presence of an adequate magnetic field will cause the output $\mathrm{V}_{\mathrm{Q}}$ to switch to a High level (i.e. off state ).

| Type | Marking | Ordering Code | Package |
| :--- | :--- | :--- | :--- |
| TLE 4917 | 17s | Q62705K 605 | P-TSOP6-6-2 |

## Pin Configuration

(top view)


Figure 1

Pin Definitions and Functions

| Pin | Symbol | Function |
| :--- | :--- | :--- |
| 1 | $V_{\mathrm{S}}$ | Supply Voltage |
| 2 | $G N D$ | Ground |
| 3 | $Q$ | Open Drain Input |
| 4 | GND | Ground |
| 5 | $G N D$ | Ground |
| 6 | $P R G$ | Programming Input |



Figure 2 Block Diagram

## Circuit Description

The Low Power Hall IC Switch comprises a Hall probe, bias generator, compensation circuits, oscillator, output latch and an n-channel open drain output transistor.

The bias generator provides currents for the Hall probe and the active circuits. Compensation circuits stabilize the temperature behavior and reduce technology variations.

The Active Error Compensation rejects offsets in signal stages and the influence of mechanical stress to the Hall probe caused by molding and soldering processes and other thermal stresses in the package. This chopper technique together with the threshold generator and the comparator ensures high accurate magnetic switching points.

Very low power consumption is achieved with a timing scheme controlled by an oscillator and a sequencer. This circuitry activates the sensor for $50 \mu$ (typical operating time) sets the output state after sequential questioning of the switch points and latches it with the beginning of the following standby phase (typ. 130 ms ). In the standby phase the average current is reduced to typical $3.5 \mu \mathrm{~A}$. Because of the long standby time compared to the operating time the overall averaged current is only slightly higher than the standby current.

By connecting the programming pin to GND (normal to $V_{S}$ ) the Output State can be inverted to further reduce the current consumption in applications where a high magnetic field is the
normal state. In that case the output $Q$ is off at high magnetic fields and no current is flowing in the open drain transistor.

The output transistor can sink up to 1 mA with a maximal saturation voltage $V_{\text {QSAT }}$.

## Absolute Maximum Ratings

| Parameter | Symbol | Limit Values |  | Unit | Notes |
| :--- | :--- | :--- | :--- | :--- | :--- |
|  |  | min. | max. |  |  |
| Supply Voltage | $V_{\mathrm{S}}$ | -0.3 | 5.5 | V |  |
| Supply Current | $I_{\mathrm{S}}$ | -1 | 2.5 | mA |  |
| Output Voltage | $V_{\mathrm{Q}}$ | -0.3 | 5.5 | V |  |
| Output Current | $I_{\mathrm{Q}}$ | -1 | 2 | mA |  |
| Programming Pin Voltage | $V_{\mathrm{PRG}}$ | -0.3 | $5.5^{1)}$ | V |  |
| Junction temperature | $T_{\mathrm{J}}$ | -40 | 150 | ${ }^{\circ} \mathrm{C}$ |  |
| Storage temperature | $T_{\mathrm{S}}$ | -40 | 150 | ${ }^{\circ} \mathrm{C}$ |  |
| Magnetic Flux Density | $B$ | - | unlimited | mT |  |
| Thermal Resistance <br> P-TSOP6-6-2 | $R_{\mathrm{th} \text { JA }}$ | - | 35 | $\mathrm{~K} / \mathrm{W}$ |  |

${ }^{1)} V_{\text {PRG }}$ must not exceed $V s$ by more than 0.3 V
Note: Stresses above those listed here may cause permanent damage to the device. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ESD Protection

Human Body Model (HBM) tests according to:
EOS/ESD Association Standard S5.1-1993 and Mil. Std. 883D method 3015.7

| Parameter | Symbol | Limit Values |  | Unit | Notes |
| :--- | :--- | :--- | :--- | :--- | :--- |
|  |  | Min. | max. |  |  |
| ESD Voltage | $V_{\text {ESD }}$ | $\pm 2$ |  | kV | $R=1.5 \mathrm{k} \Omega$, |
|  |  |  |  |  | $C=100 \mathrm{pF} ;$ |
|  |  |  |  |  | $T=25^{\circ} \mathrm{C}$ |

Operating Range

|  | Symbol | Limit Values |  |  | Unit | Notes |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  | Min. | typ. | max. |  |  |
| Supply voltage | $V_{\mathrm{S}}$ | 2.4 | 2.7 | 5.5 | V | ${ }^{1}$ |
| Output voltage | $V_{\mathrm{Q}}$ | -0.3 | 2.7 | 5.5 | V |  |
| Programming Pin Voltage | $V_{\mathrm{PRG}}$ | -0.3 | 0 | 0.3 | V | Inverted output <br> state |
|  |  | $V_{\mathrm{S}}-$ <br> 0.3 | $V_{\mathrm{S}}$ | $V_{\mathrm{S}}+$ <br> 0.3 |  | Standard output <br> state |
| Ambient Temperature | $T_{\mathrm{A}}$ | -40 | 25 | 85 | ${ }^{\circ} \mathrm{C}$ |  |

${ }^{1)} \mathrm{A}$ Ceramic Bypass Capacitor of 10 nF at $V_{\mathrm{S}}$ to GND is highly recommended.

## AC/DC Characteristics

| Parameter | Symbol | Limit Values |  |  | Unit | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min. | typ. | Max. |  |  |
| Averaged Supply Current | $I_{\text {SAVG }}$ | 1 | 4 | 20 | $\mu \mathrm{A}$ |  |
| Averaged Supply Current during Operating Time | $I_{\text {SOPAVG }}$ | 0.5 | 1.1 | 2.5 | mA |  |
| Transient Peak Supply Current during Operating Time | $I_{\text {SOPT }}$ | - | - | 2.5 | mA | $t<100 \mathrm{~ns}$ |
| Supply Current during Standby Time | $I_{\text {SSTB }}$ | 1 | 3.5 | 20 | $\mu \mathrm{A}$ |  |
| Output Saturation Voltage | $V_{\text {QSAT }}$ | - | 0.13 | 0.4 | V | $\mathrm{I}_{\mathrm{Q}}=1 \mathrm{~mA}$ |
| Output Leakage Current | $I_{\text {QLEAK }}$ | - | 0.01 | 1 | $\mu \mathrm{A}$ |  |
| Output Rise Time | $t_{r}$ | - | 0.3 | 1 | $\mu \mathrm{s}$ | $\begin{aligned} & R_{\mathrm{L}}=2.7 \mathrm{k} \Omega ; \\ & C_{\mathrm{L}}=10 \mathrm{pF} \\ & \hline \end{aligned}$ |
| Output Fall Time | $t_{\text {f }}$ | - | 0.1 | 1 | $\mu \mathrm{s}$ | $\begin{aligned} & R_{\mathrm{L}}=2.7 \mathrm{k} \Omega ; \\ & C_{\mathrm{L}}=10 \mathrm{pF} \end{aligned}$ |
| Operating Time | $t_{\text {op }}$ | 15 | 50 | $93^{1 / 2)}$ | $\mu \mathrm{S}$ |  |
| Standby Time | $t_{\text {stb }}$ | - | 130 | $240{ }^{31}$ | ms |  |
| Duty Cycle | $t_{\text {op }} / t_{\text {stb }}$ | - | 0.039 | - | \% |  |
| Start-up Time of IC | $t_{\text {stu }}$ | - | 6 | 12 | $\mu \mathrm{s}$ | ${ }^{4)}$ |

${ }^{1)}$ for $V_{\mathrm{S}}=3.5 \mathrm{~V}$ the max. Operating Time $t_{\mathrm{op} \max }=85 \mu \mathrm{~s}$
${ }^{2)}$ includes the Start-up Time $t_{\text {stu }}$
${ }^{3)}$ for $V_{S}=3.5 \mathrm{~V}$ the max. Standby Time $t_{\text {stb max }}=220 \mathrm{~ms}$
${ }^{4)}$ initial power on time. $V_{\mathrm{S}}$ must be applied in this time ( typ. $6 \mu \mathrm{~s}$ to max. $12 \mu \mathrm{~s}$ ) to get already a valid output state after the first operating phase (typ. $56 \mu \mathrm{~s}$ ). For rise times of $\mathrm{V}_{\mathrm{S}}>12 \mu \mathrm{~s}$, the output state is valid after the second operating phase (includes one standby phase), e.g. happens only when the battery in flip phones is changed.

## Magnetic Characteristics

PRG Pin Connected to $V_{s}$

| Parameter | Symbol | Limit Values |  |  |  | Unit |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Notes |  |  |  |  |  |  |
| Operate Points |  | Min. | typ. | max. |  |  |
|  | B OPS | 3.5 | 5 | 7 | mT | 11 |
|  | $\mathrm{~B}_{\text {OPN }}$ | -7 | -5 | -3.5 | mT |  |
| Release Points | $\mathrm{B}_{\text {RPS }}$ | 2.2 | 4 | 6 | mT | 11 |
|  | $\mathrm{~B}_{\text {RPN }}$ | -6 | -4 | -2.2 | mT |  |
| Hysteresis | $\mathrm{B}_{\text {HYS }}$ | 0.2 | 1 | 2 | mT |  |
| 1) Positive magnetic fields |  |  |  |  |  |  |

PRG Pin Connected to GND

| Parameter | Symbol | Limit Values |  |  | Unit | Notes |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  | Min. | typ. | max. |  |  |
| Operate Points | B OPS | 2.2 | 4 | 6 | mT | 11 |
|  | B OPN | -6 | -4 | -2.2 | mT |  |
| Release Points | $\mathrm{B}_{\text {RPS }}$ | 3.5 | 5 | 7 | mT | 11 |
|  | $\mathrm{~B}_{\text {RPN }}$ | -7 | -5 | -3.5 | mT |  |
| Hysteresis | $\mathrm{B}_{\text {HY }}$ | 0.2 | 1 | 2 | mT |  |

${ }^{1)}$ Positive magnetic fields are related to the approach of a magnetic south pole to the branded side of package

Note: The listed AC/DC and magnetic characteristics are ensured over the operating range of the integrated circuit. Typical characteristics specify mean values expected over the production spread. If not other specified, typical characteristics apply at $T_{j}=25{ }^{\circ} \mathrm{C}$ and $V_{S}=2.7 \mathrm{~V}$.


Figure 3 Timing Diagram


Figure 4 Programming of Output with the PRG Pin

All curves reflect typical values at the given parameters for $T_{A}$ in ${ }^{\circ} \mathrm{C}$ and $\mathrm{V}_{\mathrm{S}}$ in V .

Magnetic Switching Points versus Temperature ( $\mathrm{V}_{\mathrm{S}}=2.7 \mathrm{~V}$ ) (PRG Pin Connected to $\mathbf{V}_{\mathbf{s}}$ )


Supply current Isopavg during Operating Time versus Temperature ( $\mathrm{V}_{\mathrm{s}}=2.7 \mathrm{~V}$ )

Magnetic Switching Points versus
Supply Voltage $\mathrm{V}_{\mathrm{S}}\left(\mathrm{T}_{\mathrm{A}}=20^{\circ} \mathrm{C}\right)$
(PRG Pin Connected to $\mathrm{V}_{\mathrm{s}}$ )


Supply current Isopavg during Operating Time versus Supply Voltage $\mathrm{V}_{\mathrm{S}}\left(\mathrm{T}_{\mathrm{A}}=20^{\circ} \mathrm{C}\right)$


Supply current Isstв $^{\text {during Standby }}$ Time versus Temperature ( $\mathrm{V}_{\mathrm{S}}=2.7 \mathrm{~V}$ )


Output Saturation voltage $\mathrm{V}_{\text {QSAT }}$ versus Temperature ( $\mathrm{I}_{\mathrm{Q}}=1 \mathrm{~mA}$ )


Supply current Isstв $^{\text {during Standby }}$ Time versus Supply Voltage $\mathrm{V}_{\mathrm{S}}\left(\mathrm{T}_{\mathrm{A}}=20^{\circ} \mathrm{C}\right)$


Standby Time $\mathrm{t}_{\text {stb }}$ versus Temperature ( $\mathrm{V}_{\mathrm{S}}=\mathbf{2 . 7 V}$ )



Figure 5 Marking and Tape Loading Orientation


Figure 6 Foot Print Reflow Soldering

## Package Dimensions

## P-TSOP6-6-2 (Plastic Thin Small Outline-Package)


$\square$ (6)


GPX09300
weight : 0.015 g
coplanary : 0.1 mm

## Sorts of Packing

Package outlines for tubes, trays etc. are contained in our
Data Book "Package Information".
SMD = Surface Mounted Device
Dimensions in mm


## Application circuit TLE 4917

The minimum value for the pull up resistor can be calculated with the power supply voltage Vs , the maximum current $\mathrm{I}_{\mathrm{Qmax}}$ and the minimum output saturation voltage $\mathrm{V}_{\mathrm{QSAT}}$.

## Example:

for $V s=3 \mathrm{~V}$ : $R_{\text {Lmin }}=\left(\mathrm{Vs}-\mathrm{V}_{\text {QSAT }} m i n\right) / I Q \max =(3 \mathrm{~V}-0,1 \mathrm{~V}) / 0,002 \mathrm{~A}=1435 \Omega$
Larger values for $R_{L}$ will reduce the current $I_{Q}$ and therefore the power consumption. If the resistor RL is very large (>100 $\mathrm{k} \Omega$ ) a capacitor (app. 10pF) between Output and GND pin could be useful if capacitive coupled noise occurs.
The load at the output $Q$ should have a large input resistance to reduce the current trough $R_{L}$ and the power consumption.
The TLE 4917 has 3 ground pins. From a mechanical point of view all ground pins should be connected to ground. Shortest wires should be used to avoid ground loops.
If there is a need to reduce the number of used ground-pins any ground-pin combination may me used. Furthermore it is possible using only one ground-pin at the application, all pins are equivalent.
The capacitor C is highly recommended to reduce noise on the power supply voltage and it will improve the EMI/EMC performance.
Furthermore it decreases the transient peak supply current during operation time. The IC toggles between low and high current consumption. This behaviour might produce additional noise at the power supply. The capacitor will reduce this noise.
Furthermore this capacitor is used to supply the sensor if microbreaks (short loss of supply voltage) occur.

Shortest connection wires between IC and capacitor should be used to avoid noise.
The switch S1 shows the programming feature of the output.

## Example:

If the PRG-pin is connected to Vs the IC will hold the output $Q$ at a high voltage level for $B=0 \mathrm{mT}$ in this circuit. A magnetic field larger than the operating point will switch the output to low level. In typical applications the PRG-pin is connected directly to Vs or to GND depending on the technical needs. Avoid using a floating PRG-pin.
technologies

| TLE 4917 |  |
| :--- | :--- |
| Revision History: $\quad 2002-08-22$ |  |
| Previous Version: |  |
| Page | Subjects (major changes since last revision) |
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