

# Ultra Low Quiescent Current Linear Voltage Regulator

**TLS810A1** 

TLS810A1LDV50

# Linear Voltage Regulator

# **Data Sheet**

Rev. 1.0, 2016-03-15

**Automotive Power** 



TLS810A1 TLS810A1LDV50





# 1 Overview

### **Features**

- Ultra Low Quiescent Current of 5 μA
- Wide Input Voltage Range of 2.75 V to 42 V
- Output Current Capacity up to 100 mA
- Low Drop Out Voltage of typ. 200 mV @ 100 mA
- Output Current Limit Protection
- Overtemperature Shutdown
- Available in PG-TSON-10 Package
- Wide Temperature Range
- Green Product (RoHS Compliant)
- AEC Qualified



Figure 1 PG-TSON-10

Туре	Package	Marking
TLS810A1LDV50	PG-TSON-10	810A1V5

#### TLS810A1LDV50



#### Overview

#### Description

The TLS810A1 is a linear voltage regulator featuring wide input voltage range, low drop out voltage and ultra low quiescent current.

With an input voltage range of 2.75 V to 42 V and ultra low quiescent of only 5  $\mu$ A, the regulator is perfectly suitable for automotive or any other supply systems connected permanently to the battery.

The TLS810A1LDV50 is the fixed 5 V output version with an accuracy of 2 % and output current capability up to 100 mA.

The new regulation concept implemented in TLS810A1 combines fast regulation and very good stability while requiring only a small ceramic capacitor of 1  $\mu$ F at the output.

The tracking region starts already at input voltages of 2.75 V (extended operating range). This makes the TLS810A1 also suitable to supply automotive systems that need to operate during cranking condition.

Internal protection features like output current limitation and overtemperature shutdown are implemented to protect the device against immediate damage due to failures like output short circuit to GND, over-current and over-temperature.

### **Choosing External Components**

An input capacitor  $C_{\rm l}$  is recommended to compensate line influences. The output capacitor  $C_{\rm Q}$  is necessary for the stability of the regulating circuit. Stability is guaranteed at values  $C_{\rm Q} \ge 1~\mu{\rm F}$  and an ESR  $\le 100~\Omega$  within the whole operating range.



**Block Diagram** 

# 2 Block Diagram

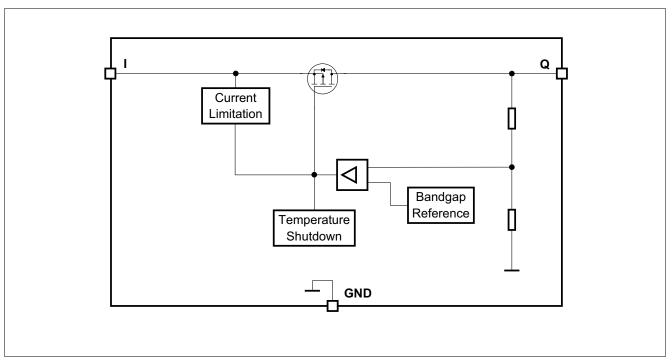


Figure 2 Block Diagram TLS810A1



**Pin Configuration** 

# **3** Pin Configuration

# 3.1 Pin Assignment in PG-TSON-10 Package

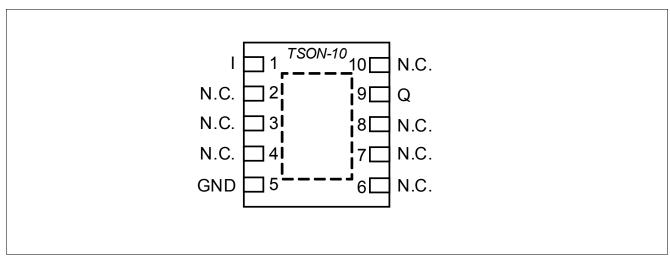


Figure 3 Pin Configuration TLS805A1TLS810A1 in PG-TSON-10 package

# 3.2 Pin Definitions and Functions in PG-TSON-10 Package

Pin	Symbol	Function
1	1	Input It is recommended to place a small ceramic capacitor (e.g. 100 nF) to GND, close
		to the IC terminals, in order to compensate line influences.
2	N.C.	Not connected
3	N.C.	Not connected
4	N.C.	Not connected
5	GND	Ground
6	N.C.	Not connected
7	N.C.	Not connected
8	N.C.	Not connected
9	Q	Output
		Connect an output capacitor $C_Q$ to GND close to the IC's terminals, respecting the values specified for its capacitance and ESR in <b>Table 2 "Functional Range" on</b>
		Page 8.

# **TLS810A1LDV50**



# **Pin Configuration**

Pin	Symbol	Function
10	N.C.	Not connected
Pad	_	Exposed Pad
		Connect to heatsink area.
		Connect to GND.



#### **General Product Characteristics**

# 4 General Product Characteristics

# 4.1 Absolute Maximum Ratings

# Table 1 Absolute Maximum Ratings<sup>1)</sup>

 $T_i$  = -40 °C to +150 °C; all voltages with respect to ground (unless otherwise specified)

Parameter	Symbol	Symbol Value:			Unit	Note or	Number
		Min.	Тур.	Max.		<b>Test Condition</b>	
Voltage Input I	•		, ,	, ,	<u>'</u>	,	
Voltage	$V_{I}$	-0.3	_	45	V	_	P_4.1.1
Voltage Output Q	•	<del></del>			<del>-</del>	•	*
Voltage	$V_{Q}$	-0.3	-	7	V	_	P_4.1.2
Temperatures			, ,	, ,	<u>'</u>	,	
Junction Temperature	$T_{\rm j}$	-40	_	150	°C	_	P_4.1.3
Storage Temperature	$T_{\rm stg}$	-55	_	150	°C	-	P_4.1.4
ESD Absorption			, ,	, ,	<u>'</u>	,	
ESD Susceptibility to GND	$V_{\rm ESD,HBM}$	-2	-	2	kV	HBM <sup>2)</sup>	P_4.1.5
ESD Susceptibility to GND	$V_{\rm ESD,CDM}$	-750	_	750	V	CDM <sup>3)</sup> at all pins	P_4.1.6

<sup>1)</sup> Not subject to production testing, specified by design.

#### **Notes**

- 1. Stresses above the ones listed here may cause permanent damage to the device. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.
- 2. Integrated protection functions are designed to prevent IC destruction under fault conditions described in the data sheet. Fault conditions are considered as "outside" normal operating range. Protection functions are not designed for continuous repetitive operation.

<sup>2)</sup> ESD susceptibility, HBM according to ANSI/ESDA/JEDEC JS001 (1.5 k $\Omega$ , 100 pF)

<sup>3)</sup> ESD susceptibility, Charged Device Model "CDM" according JEDEC JESD22-C101



#### **General Product Characteristics**

# 4.2 Functional Range

Table 2 Functional Range

Parameter	Symbol		Values	;	Unit	Note or Test Condition	Number
		Min.	Тур.	Max.			
Input Voltage Range	$V_{I}$	$V_{\rm Q,nom} + V_{\rm dr}$	_	42	٧	_1)	P_4.2.1
Extended Input Voltage Range	$V_{\rm l,ext}$	2.75	_	42	V	_2)	P_4.2.2
Output Capacitor	$C_{Q}$	1	_	_	μF	_3)4)	P_4.2.3
Output Capacitor's ESR	$ESR(C_Q)$	_	_	100	Ω	_4)	P_4.2.4
Junction temperature	T <sub>i</sub>	-40	_	150	°C	_	P_4.2.5

<sup>1)</sup> Output current is limited internally and depends on the input voltage, see Electrical Characteristics for more details.

Note: Within the functional or operating range, the IC operates as described in the circuit description. The electrical characteristics are specified within the conditions given in the Electrical Characteristics table.

<sup>2)</sup> When  $V_{l}$  is between  $V_{l,ext.min}$  and  $V_{Q,nom} + V_{dr}$ ,  $V_{Q} = V_{l} - V_{dr}$ . When  $V_{l}$  is below  $V_{l,ext,min}$ ,  $V_{Q}$  can drop down to 0 V.

<sup>3)</sup> The minimum output capacitance requirement is applicable for a worst case capacitance tolerance of 30%.

<sup>4)</sup> Not subject to production testing, specified by design.



#### **General Product Characteristics**

### 4.3 Thermal Resistance

Note: This thermal data was generated in accordance with JEDEC JESD51 standards. For more information, go to www.jedec.org.

Table 3 Thermal Resistance TLS810A1 in PG-TSON-10 Package

Parameter	Symbol		Value	s	Unit	Note or Test Condition	Number
		Min.	Тур.	Max.			
Package Version		1					
Junction to Case <sup>1)</sup>	$R_{thJC}$	-	13	_	K/W	-	P_4.3.1
Junction to Ambient <sup>1)</sup>	$R_{thJA}$	-	60	_	K/W	2s2p board <sup>2)</sup>	P_4.3.2
Junction to Ambient <sup>1)</sup>	$R_{thJA}$	-	184	-	K/W	1s0p board, footprint only <sup>3)</sup>	P_4.3.3
Junction to Ambient <sup>1)</sup>	$R_{thJA}$	-	75	-	K/W	1s0p board, 300 mm <sup>2</sup> heatsink area on PCB <sup>3)</sup>	P_4.3.4
Junction to Ambient <sup>1)</sup>	$R_{thJA}$	-	64	-	K/W	1s0p board, 600 mm <sup>2</sup> heatsink area on PCB <sup>3)</sup>	P_4.3.5

<sup>1)</sup> Not subject to production test, specified by design

<sup>2)</sup> Specified  $R_{thJA}$  value is according to Jedec JESD51-2,-5,-7 at natural convection on FR4 2s2p board; The Product (Chip+Package) was simulated on a 76.2 x 114.3 x 1.5 mm<sup>3</sup> board with 2 inner copper layers (2 x 70 $\mu$ m Cu, 2 x 35 $\mu$ m Cu). Where applicable a thermal via array under the exposed pad contacted the first inner copper layer.

<sup>3)</sup> Specified  $R_{thJA}$  value is according to JEDEC JESD 51-3 at natural convection on FR4 1s0p board; The Product (Chip+Package) was simulated on a 76.2 × 114.3 × 1.5 mm<sup>3</sup> board with 1 copper layer (1 x 70 $\mu$ m Cu).



# 5 Block Description and Electrical Characteristics

# 5.1 Voltage Regulation

The output voltage  $V_Q$  is divided by a resistor network. This fractional voltage is compared to an internal voltage reference and the pass transistor is driven accordingly.

The control loop stability depends on the output capacitor  $C_Q$ , the load current, the chip temperature and the internal circuit structure. To ensure stable operation, the output capacitor's capacitance and its equivalent series resistor ESR requirements given in "Functional Range" on Page 8 have to be maintained. For details see the typical performance graph Output Capacitor Series Resistor ESR( $C_Q$ ) versus Output Current  $C_Q$ . Since the output capacitor is used to buffer load steps, it should be sized according to the application's needs.

An input capacitor  $C_1$  is not required for stability, but is recommended to compensate line fluctuations. An additional reverse polarity protection diode and a combination of several capacitors for filtering should be used, in case the input is connected directly to the battery line. Connect the capacitors close to the regulator terminals.

In order to prevent overshoots during start-up, a smooth ramping up function is implemented. This ensures almost no overshoots during start-up, mostly independent from load and output capacitance.

Whenever the load current exceeds the specified limit, e.g. in case of a short circuit, the output current is limited and the output voltage decreases.

The overtemperature shutdown circuit prevents the IC from immediate destruction under fault conditions (e.g. output continuously short-circuit) by switching off the power stage. After the chip has cooled down, the regulator restarts. This oscillatory thermal behaviour causes the junction temperature to exceed the maximum rating of 150°C and can significantly reduce the IC's lifetime.

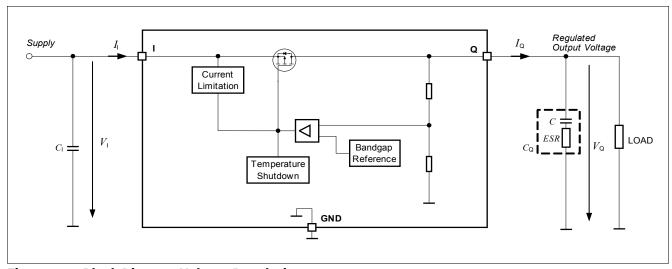


Figure 4 Block Diagram Voltage Regulation



#### **Table 4 Electrical Characteristics**

 $T_{\rm j}$  = -40 °C to +150 °C,  $V_{\rm j}$  = 13.5 V, all voltages with respect to ground (unless otherwise specified). Typical values are given at  $T_{\rm i}$  = 25 °C,  $V_{\rm i}$  = 13.5 V.

Parameter	Symbol	Values			Unit	<b>Note or Test Condition</b>	Number
		Min.	Тур.	Max.			
Output Voltage Precision	V <sub>Q</sub>	4.90	5.00	5.10	V	50 μA $\leq I_Q \leq$ 100 mA, 5.7 V $\leq V_1 \leq$ 28 V	P_5.1.1
Output Voltage Precision	$V_{Q}$	4.90	5.00	5.10	V	$50 \mu A \le I_Q \le 50 mA$ , $5.7 V \le V_1 \le 42 V$	P_5.1.2
Output Current Limitation	$I_{\mathrm{Q,lim}}$	110	190	260	mA	$0 \text{ V} \le V_{\text{Q}} \le V_{\text{Q,nom}} - 0.1 \text{ V}$	P_5.1.3
Line Regulation steady-state	$\Delta V_{ m Q,line}$	-	1	20	mV	$I_Q = 1 \text{ mA}, 6 \text{ V} \le V_1 \le 32 \text{ V}$	P_5.1.4
Load Regulation steady-state	$\Delta V_{ m Q,load}$	-20	-1	-	mV	$V_1 = 6 \text{ V},$ 50 $\mu \text{A} \le I_Q \le 100 \text{ mA}$	P_5.1.5
Dropout Voltage <sup>1)</sup> $V_{dr} = V_{l} - V_{Q}$	$V_{\rm dr}$	-	200	550	mV	$I_{\rm Q} = 100  {\rm mA}$	P_5.1.6
Ripple Rejection <sup>2)</sup>	PSRR	-	55	-	dB	$I_{\rm Q}$ = 50 mA, $f_{\rm ripple}$ = 100 Hz, $V_{\rm ripple}$ = 0.5 $V_{\rm p-p}$	P_5.1.7
Overtemperature Shutdown Threshold <sup>2)</sup>	$T_{\rm j,sd}$	151	175	-	°C	T <sub>j</sub> increasing	P_5.1.8
Overtemperature Shutdown Threshold Hysteresis <sup>2)</sup>	$T_{\rm j,sdh}$	-	10	-	K	T <sub>j</sub> decreasing	P_5.1.9

<sup>1)</sup> Measured when the output voltage  $V_Q$  has dropped 100 mV from the nominal value obtained at  $V_I = 13.5$ V

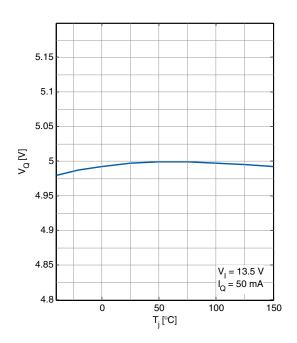
<sup>2)</sup> Not subject to production test, specified by design



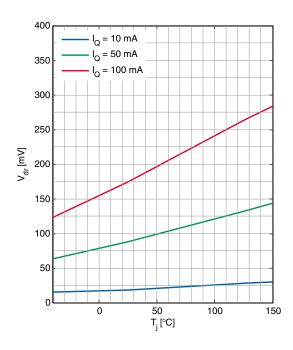
# 5.2 Typical Performance Characteristics Voltage Regulation

### **Typical Performance Characteristics**

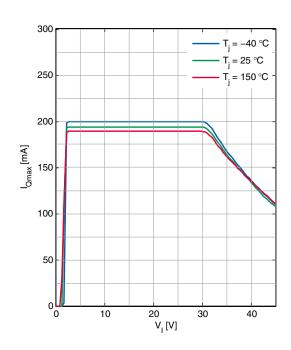
# Output Voltage $V_{\mathbf{Q}}$ versus Junction Temperature $T_{\mathbf{i}}$



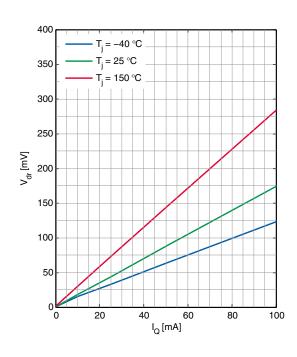
# Dropout Voltage $V_{dr}$ versus Junction Temperature $T_{i}$



# Output Current $I_{\mathbf{Q}}$ versus Input Voltage $V_{\mathbf{I}}$

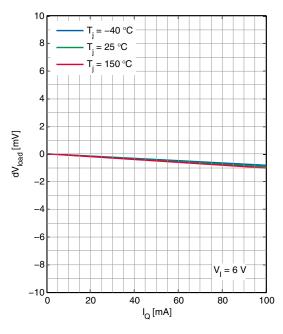


Dropout Voltage  $V_{\rm dr}$  versus Output Current  $I_{\rm Q}$ 

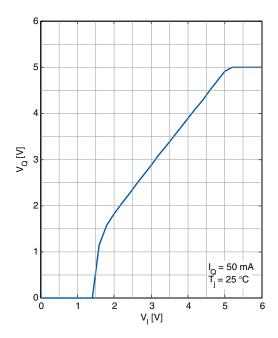




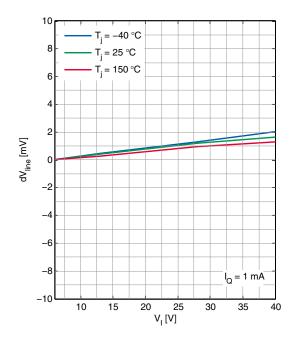
# Load Regulation $\Delta V_{\rm Q,load}$ versus Output Current $I_{\rm O}$



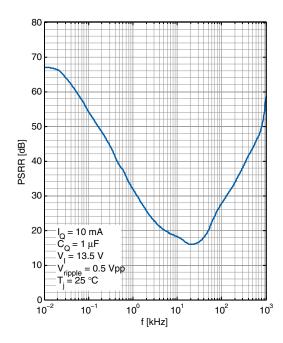
Output Voltage  $V_{\mathbf{Q}}$  versus Input Voltage  $V_{\mathbf{I}}$ 



# Line Regulation $\Delta V_{\rm Q,line}$ versus Input Voltage $V_{\rm I}$

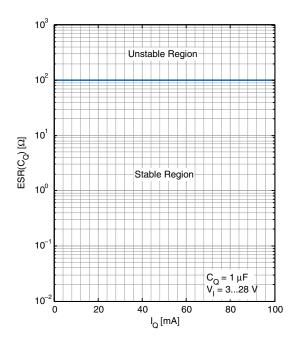


Power Supply Ripple Rejection PSRR versus Ripple Frequency  $f_{\mathbf{r}}$ 





# Output Capacitor Series Resistor $ESR(C_{\mathbb{Q}})$ versus Output Current $I_{\mathbb{Q}}$





# **5.3** Current Consumption

# **Table 5 Electrical Characteristics Current Consumption**

 $T_{\rm i}$  = -40 °C to +150 °C,  $V_{\rm i}$  = 13.5 V (unless otherwise specified).

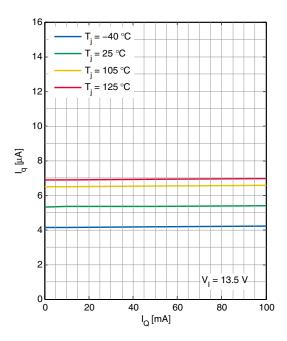
Parameter	Symbol	Values		Unit	Note or Test Condition	Number	
		Min.	Тур.	Max.			
Current Consumption $I_q = I_1 - I_Q$	I <sub>q</sub>	_	5	7.5	μΑ	$I_{\rm Q} = 50 \ \mu \text{A}, T_{\rm j} = 25 \ ^{\circ}\text{C}$	P_5.2.1
Current Consumption $I_q = I_1 - I_Q$	I <sub>q</sub>	_	6.5	10	μΑ	$I_{\rm Q} = 50 \ \mu \text{A}, T_{\rm j} < 105 \ ^{\circ}\text{C}$	P_5.2.2
Current Consumption $I_q = I_I - I_Q$	I <sub>q</sub>	_	7	11	μΑ	$I_{\rm Q} = 50 \ \mu \text{A}, T_{\rm j} < 125 ^{\circ}\text{C}$	P_5.2.3
Current Consumption $I_q = I_1 - I_Q$	I <sub>q</sub>	_	7	11	μΑ	$I_{\rm Q}$ = 100 mA, $T_{\rm j}$ < 125 °C	P_5.2.4



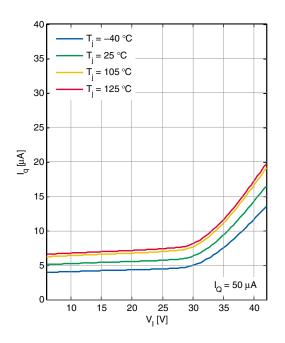
# **5.4** Typical Performance Characteristics Current Consumption

# **Typical Performance Characteristics**

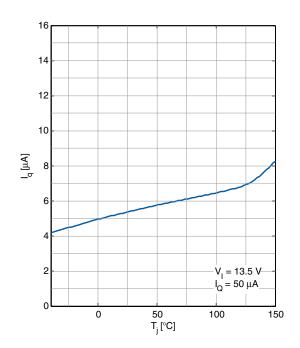
# Current Consumption $I_{\bf q}$ versus Output Current $I_{\bf O}$



# Current Consumption $I_{\bf q}$ versus Input Voltage $V_{\bf l}$



# Current Consumption $I_{\bf q}$ versus Junction Temperature $T_{\bf i}$





#### **Application Information**

# **6** Application Information

Note: The following information is given as a hint for the implementation of the device only and shall not be regarded as a description or warranty of a certain functionality, condition or quality of the device.

# 6.1 Application Diagram

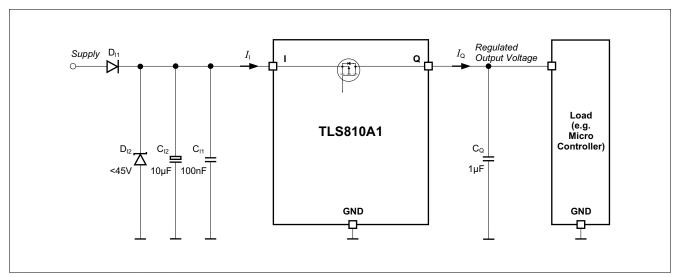


Figure 5 Application Diagram

# **6.2** Selection of External Components

### 6.2.1 Input Pin

The typical input circuitry for a linear voltage regulator is shown in the application diagram above.

A ceramic capacitor at the input, in the range of 100 nF to 470 nF, is recommended to filter out the high frequency disturbances imposed by the line e.g. ISO pulses 3a/b. This capacitor must be placed very close to the input pin of the linear voltage regulator on the PCB.

An aluminum electrolytic capacitor in the range of  $10\,\mu\text{F}$  to  $470\,\mu\text{F}$  is recommended as an input buffer to smooth out high energy pulses, such as ISO pulse 2a. This capacitor should be placed close to the input pin of the linear voltage regulator on the PCB.

An overvoltage suppressor diode can be used to further suppress any high voltage beyond the maximum rating of the linear voltage regulator and protect the device against any damage due to over-voltage.

The external components at the input are not mandatory for the operation of the voltage regulator, but they are recommended in case of possible external disturbances.



#### **Application Information**

# 6.2.2 Output Pin

An output capacitor is mandatory for the stability of linear voltage regulators.

The requirement to the output capacitor is given in "Functional Range" on Page 8. The graph "Output Capacitor Series Resistor ESR( $C_Q$ ) versus Output Current  $I_Q$ " on Page 14 shows the stable operation range of the device.

TLS810A1 is designed to be stable with extremely low ESR capacitors. According to the automotive environment, ceramic capacitors with X5R or X7R dielectrics are recommended.

The output capacitor should be placed as close as possible to the regulator's output and GND pins and on the same side of the PCB as the regulator itself.

In case of rapid transients of input voltage or load current, the capacitance should be dimensioned in accordance and verified in the real application that the output stability requirements are fulfilled.

#### 6.3 Thermal Considerations

Knowing the input voltage, the output voltage and the load profile of the application, the total power dissipation can be calculated:

$$P_{D} = (V_{I} - V_{Q}) \times I_{Q} + V_{I} \times I_{q}$$
(6.1)

with

- P<sub>D</sub>: continuous power dissipation
- V<sub>i</sub>: input voltage
- V<sub>o</sub>: output voltage
- I<sub>o</sub>: output current
- I<sub>a</sub>: quiescent current

The maximum acceptable thermal resistance  $R_{\rm thJA}$  can then be calculated:

$$R_{thJA, max} = \frac{T_{j, max} - T_{a}}{P_{D}}$$
(6.2)

with

- $T_{i,max}$ : maximum allowed junction temperature
- T<sub>a</sub>: ambient temperature

Based on the above calculation the proper PCB type and the necessary heat sink area can be determined with reference to the specification in "Thermal Resistance" on Page 9.

#### **Example**

Application conditions:

$$V_1 = 13.5 \text{ V}$$

$$V_0 = 5 \text{ V}$$

$$I_0 = 80 \text{ mA}$$

$$T_a = 105 \, ^{\circ}\text{C}$$



#### **Application Information**

```
Calculation of R_{\text{thJA,max}}:

P_{\text{D}} = (V_{\text{I}} - V_{\text{Q}}) \times I_{\text{Q}} + V_{\text{I}} \times I_{\text{q}}

= (13.5 \text{V} - 5 \text{V}) \times 80 \text{ mA} + 13.5 \text{ V} \times 0.016 \text{ mA}

= 0.68 \text{ W}

R_{\text{thJA,max}} = (T_{\text{j,max}} - T_{\text{a}}) / P_{\text{D}}

= (150 \, ^{\circ}\text{C} - 105 \, ^{\circ}\text{C}) / 0.68 \text{ W}

= 66.2 \text{ K/W}
```

As a result, the PCB design must ensure a thermal resistance  $R_{\rm thJA}$  lower than 66.2 K/W. According to "Thermal Resistance" on Page 9, at least 600 mm<sup>2</sup> heatsink area is needed on the FR4 1s0p PCB, or the FR4 2s2p board can be used.

# 6.4 Reverse Polarity Protection

TLS810A1 is not self protected against reverse polarity faults. To protect the device against negative supply voltage, an external reverse polarity diode is needed, as shown in **Figure 5**. The absolute maximum ratings of the device as specified in "**Absolute Maximum Ratings**" on **Page 7** must be kept.

# 6.5 Further Application Information

For further information you may contact http://www.infineon.com/



# **Package Outlines**

# 7 Package Outlines

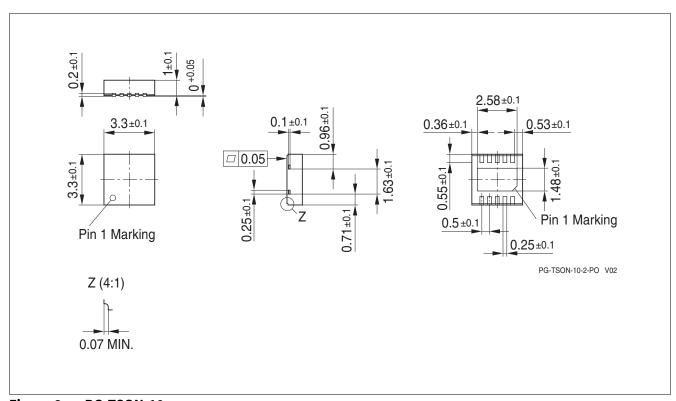


Figure 6 PG-TSON-10

# **Green Product (RoHS compliant)**

To meet the world-wide customer requirements for environmentally friendly products and to be compliant with government regulations the device is available as a green product. Green products are RoHS-Compliant (i.e Pb-free finish on leads and suitable for Pb-free soldering according to IPC/JEDEC J-STD-020).

# **TLS810A1LDV50**



# **Revision History**

# 8 Revision History

Revision	Date	Changes
1.0	2016-03-15	Datasheet - Initial version

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Edition 2016-03-15 Published by Infineon Technologies AG 81726 Munich, Germany

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