

Preliminary Specifications Subject to Change without Notice

DESCRIPTION

The JW[®]H7103 is a single channel current-limited power switch optimized for Universal Serial Bus (USB) and other hot-swap applications. The rise and fall times are controlled to minimize current overshoot or undershoot during switches on/off.

The device has fast short-circuited response time for improved overall system robustness. It provides a complete protection solution, such as reverse current blocking and limit, over-current protection, over-temperature protection and short-circuit protection, as well as controlled rise time and under-voltage lockout function. A 7.5ms de-glitch time on the open-drain flag output prevents false over-current reporting.

JWH7103 offers both DFN2X2-6 and SOT23-6 packages.

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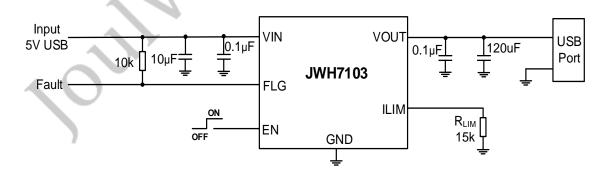
FEATURES

- Up to 1. 5A maximum load current
- ±9% accurate current limit@1.7A
- FLG: active low
- EN: active high
- Reverse current blocking during shutdown and reverse current limit during enable
- Constant-current during over-current
- Fast short-circuit response time: 2µs (typ.)
- Adjustable current limit 0.4A~1.7A
- Operating range: 2.7V 5.5V
- Over-current protection, short-circuit Protection and thermal protection
- Fault report (FAULT) with de-glitch time
- ESD protection: 2kV HBM, 750V CDM
- Available in SOT23-6 and DFN2X2-6 packages

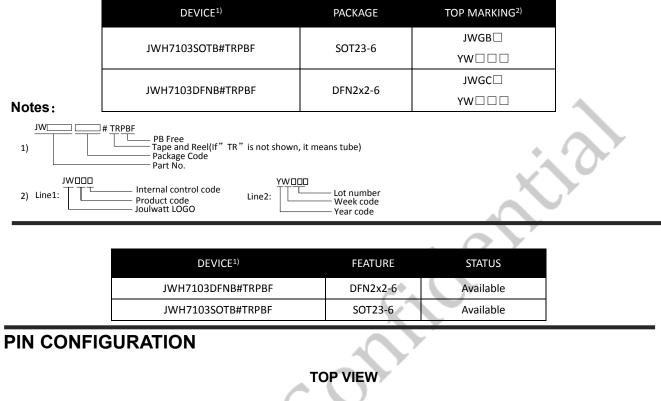
APPLICATIONS

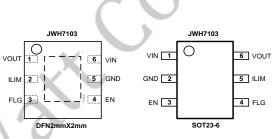
- Set-Top Boxes
- LCD TVs & Monitors
- Residential Gateways
- Laptops, Desktops, Servers, E-books, Printers, Docking
- Stations, HUBs

TYPICAL APPLICATION



ORDER INFORMATION





ABSOLUTE MAXIMUM RATING¹⁾

VIN PIN Voltage	0.3V to 7V
VOUT PIN Voltage	0.3V to 7V
Other Pins Voltage	
Junction Temperature ²⁾	
Lead Temperature	260°C
Storage Temperature	65°C to +150°C
ESD Susceptibility (Human Body Model)	±2kV
ESD Susceptibility (Charged Device Model)	±750V

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RECOMMENDED OPERATING CONDITIONS³⁾

VIN PIN Voltage	2.7V to 5.5V
VOUT PIN Voltage	0V to (VIN+0.2V)
EN/EN PIN Voltage	0V to 5.5V
High-Level Input Voltage on EN/EN	1.4V to VIN
Low-Level Input Voltage on EN/EN	0V to 0.5V
Operating Junction Temperature	40°C to 125°C

THERMAL PERFORMANCE⁴⁾

DFN2X2-6	7211.1°C/W
SOT23-6	220130°C/W

Note:

- 1) Exceeding these ratings may damage the device. These stress ratings do not imply function operation of the device at any other conditions beyond those indicated under RECOMMENDED OPERATING CONDITIONS.
- 2) The JWH7103 includes thermal protection that is intended to protect the device in overload conditions. Continuous operation over the specified absolute maximum operating junction temperature may damage the device.
- 3) The device is not guaranteed to function outside of its operating conditions.
- 4) Measured on JESD51-7, 4-layer PCB

ELECTRICAL CHARACTERISTICS

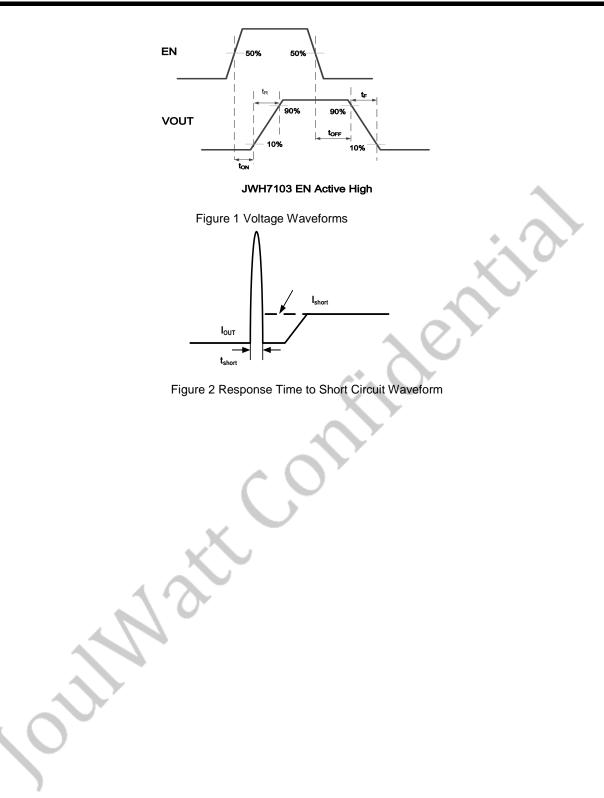
$T_{J}=-40 \ C$ to $125 \ C, 0$ unless otherwise state		nended operatii	ng conditions, $V_{IN} = 2.7V$	to 5.5V	$V_{EN} = 0$	OV or V _E	$N = V_{IN},$
ltem	Symbol		Condition ⁵⁾	Min.	Тур.	Max.	Units
Supply							
Input UVLO	V _{UVLO}		V _{IN} Rising		2.4	2.65	V
Input UVLO Hysteresis	ΔV_{uvlo}	١	/ _{IN} Decreasing		50		mV
		V_{IN} = 5.5V, Disabled, V_{OUT} = Open T _J =-40°C to 105°C			0.1	1	μA
Input Shutdown Current	SHDN	V _{IN} = 5.5V, Disabled, V _{OUT} = Open, T _J =-40°C to 125°C			0.1	2.5	μA
Input Quiescent Current	Iα	V _№ = 5.5V, Enal	bled, V_{out} = Open, R_{UM} =20k Ω	0	130	165	μA
Power Switch							
	R _{ds(ON)} –		$T_{J} = +25^{\circ}C, V_{IN} = 5.0V,$ $R_{LIM} = 15k, Load = 1A$		40	48	mΩ
Switch On Resistance		SOT23-6	T _J =-40°C to 125°C, V _{IN} = 5.0V, R _{LIM} =15k,Load=1A			60	mΩ
			$T_{J} = +25^{\circ}C, V_{IN} = 5.0V,$ $R_{LIM} = 15k, Load = 1A$		40	48	mΩ
		DFN2X2-6	T _J =-40°C to 125°C, V _{IN} = 5.0V, R _{LIM} =15k,Load=1A			60	mΩ
Output Turn-On Rise		$V_{\text{IN}}\text{=}5.5V,C_{\text{L}}\text{=}1\mu\text{F},R_{\text{LOAD}}\text{=}100\Omega.$ See Figure 1.			1.1	1.5	
Time	t _R	V_{IN} = 2.7V, C_L = 1µF, R_{LOAD} = 100 Ω .			0.7	1	ms
Output Turn-Off Fall		V _{IN} = 5.5V, C _L = 1µ	F, R_{LOAD} = 100 Ω . See Figure 1.	100		300	
Time	t⊧	V _{IN} = 2.7V, C _L = 1µ	F, R _{LOAD} = 100Ω.	100		300	μs
Enable Pin							
Turn-On Time	t _{on}	C_L = 1µF, R _L = 100Ω. See Figure 1.				3	ms
Turn-Off Time	t _{off}	C_{L} = 1µF, R _L = 100Ω. See Figure 1.				1	ms
EN High-Level Input Voltage	$V_{\text{en_H}}$			1.4			V
EN Low-Level Input Voltage	V_{EN_L}					0.7	V
Fault Flag							
FAULT Output Low Voltage	V _{ol}		I _{FAULT} = 1mA			180	mV

JWH7103

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FAULT Blanking and				_			
Latch Off	t_{Blank_OC}	Assertion or de-assertion due	5	7.5	10	ms	
Time(Over-Current)							
FAULT Blanking and							
Latch Off	$t_{Blank_{RV}}$	Assertion or de-assertion due	to reverse voltage	2	4	6	ms
Time(Reverse-Voltage)							
FAULT Off Current	FOH	$V_{FAULT} = 6V$				1	μA
Current Limit				T			
		R _{LIM} = 15kΩ	-40°C ~+125°C	1547	1700	1853	
Current-Limit Threshold		Rμμ = 20kΩ	T _J =25°C	1179	1295	1411	
(maximum DC output	ILIMIT		-40°C ~+125°C	1164	1295	1426	mA
current)		Rμμ= 50kΩ	TJ=25°C	464	520	576	
			-40°C ~+125°C	449	520	591	
Short-Circuit Fold-back	SHORT	VOUT Connected to GND	•	0	0.6		А
Current Limit	-SHOKT				0.0		
Short-Circuit Response	t _{short}	V _{IN} =5V, See Figure 2.			2		μs
Time ⁷⁾							
Reverse Voltage Protecti	on			-	-		
Reverse-Voltage	V _{RVP}	V _{OUT} -V _{IN}		90	175	230	mV
Comparator Trip Point	V RVP			30	175	200	
Reverse Current Limit	I _{ROCP}	V _{out} -V _{IN} = 200mV			0.61		А
Time from							
reverse-voltage	T _{RVP}			3	5	7	ms
condition to MOSFET	I KVP	I _{RVP}		5	5	,	1113
turn off							
Thermal Shutdown							
Thermal Shutdown					160		°C
Threshold ⁷⁾	T_{SHDN} Enabled, $R_{LOAD} = 1k\Omega$				160		°C
Thermal Shutdown							
Threshold under	T _{SHDN_OCP}	Enabled		130		°C	
Current Limit ⁷⁾	× -						
Thermal Shutdown	T _{HYS}				30		°C
Hysteresis ⁷⁾	I HYS			50		C	

Note:

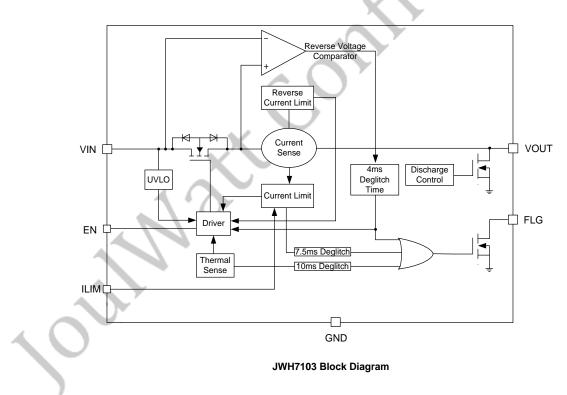
- 5) Pulse-testing techniques maintain junction temperature close to ambient temperature; thermal effects must be taken into account separately.
- 6) The discharge function is active when the device is disabled (when enable is de-asserted or during power-up power-down when VIN< VUVLO). The discharge function offers a resistive discharge path for the external storage capacitor for limited time.
- 7) Guaranteed by design



PIN DESCRIPTION

Pin SOT23-6 DFN2X2-6		Nomo	Description		
		Name			
6	1	VOUT	Output voltage		
2	5	GND	Ground(0V)		
4	3 FLG		Active-low open-drain output, asserted during over-current,		
4 5			over-temperature or reverse-voltage conditions		
3	4	EN	logic high turns on power switch.		
1	6	VIN	Input, connect a 10µF or greater ceramic capacitor from VIN to		
1			0	VIIN	GND as close to IC as possible.
5	2	ILIM	Use external resistor to set current-limit threshold;		
5		IL IIVI	Recommended $15k\Omega \le R_{LIM} \le 66.5k\Omega$.		

BLOCK DIAGRAM



FUNCTIONAL DESCRIPTION

The JWH7103 integrates high-side MOSFET optimized for Universal Serial Bus (USB) that requires protection functions. The MOSFET is driven with controlled gate voltage and slew-rate, which makes this USB device ideal for hot-swap or hot-plug applications.

FAULT Response

The Fault Flag function is realized by an open-drain circuit. The output goes active low for any of following faults: current limit threshold, short-circuit current limit, reverse voltage threshold, or thermal shutdown. In order to avoid the mis-trigger, a 7.5ms deglitch timer is inserted when an overcurrent or short circuit fault condition occurs. The FLG output remains low until over-current, short-circuit current limit, reverse current limit or over-temperature condition is removed.

Connecting a heavy capacitive load to the output of the device can cause a momentary over-current condition, which does not trigger the FAULT as long as the Fault condition lasts less than 7.5ms deglitch. This deglitch timer is also applied for over-current recovery and over-temperature recovery.

Power Supply Considerations

A local 10μ F~22 μ F/0805/10V X7R or X5R ceramic capacitor between VIN and GND, close to the device, is requested if the input supply is located more than a few inches from the device. This local capacitor can absorb the spikes on VIN pin in transient events, such as hot-plug, short-circuit, reverse-blocking. For most of conditions, 10μ F/0805/10V is highly recommended for good safety.

Additionally, bypassing the device output with a 0.1μ F to 4.7μ F ceramic capacitor improves the immunity of the device to short-circuit condition. This capacitor also prevents output from going negative during turn-off due to parasitic inductance. If the negative kick is less than -1V, a Scotty diode in parallel with VOUT pin is recommended. Otherwise, the device may go malfunction.

Generic Hot-Plug Applications

In many applications it is common to remove modules or PC boards while the main unit is still operating. These are considered hot-plug applications. Such implementations require the control of current surges. The most effective way to control the current surge is to limit and slowly ramp the current and voltage being applied to the card, similar to the Soft Start in which a power supply normally turns on. Due to the controlled rising and falling times of the switch, these devices can be used to provide a softer start-up to devices being hot-plugged into a powered system.

The UVLO feature also ensures that the switch is off after the card has been removed, and that the switch is off during the next insertion.

Under-Voltage Lockout (UVLO)

Whenever the input voltage falls below UVLO threshold (TYP. 2.4V), the power switch is turned off. This facilitates the design of hot-insertion systems where it is not possible to turn off the power switch before input power is removed.

Over-Current and Short-Circuit Protection

An internal sensing FET is employed to sense over-current conditions. Unlike current-sense resistors, sensing FETs do not increase the series resistance of the current path. When an over-current condition is detected, the switch maintains a constant output current and reduces the output voltage accordingly. Current limit will fold back to 0.6A(TYP.) when Output voltage is lower than 0.4V. Complete shutdown occurs only if the fault stays long enough to activate over-temperature protection.

Over-Current FAULT Signal

The FAULT signal will be asserted in response to OCP before the device reaches its current limit. The output current upon FAULT signal triggered will be lower than the limit value. To implement FAULT signal for precision system protection control, it is recommended to leave enough margin from maximum continuous operating current.

Over-Temperature Protection

Thermal protection prevents the IC from damage when the die temperature exceeds safe margins. This mainly occurs when heavy-overload or short-circuit faults occurs. IC implements a thermal sensing circuit to monitor the operating junction temperature. Once the die temperature rises to approximately +160°C (+130°C in case the part is under current limit), the thermal protection feature activates as follows: The internal thermal sense circuitry turns the power switch off and the FLG output is asserted, thus preventing the power damage. Once the junction switch from temperature drops to 130°C, the MOSFET restart to work.

Reverse-Current Protection

The USB specification does not allow an output device to source current back into the USB port. In a normal MOSFET switch, current will flow in reverse direction (from the output side to the input side) when the output side voltage is higher than the input side. A reverse over-current protection (ROCP) is implemented in the JWH7103 to limit reverse current. The ROCP circuit is activated when the output voltage is higher than the input voltage. After the reverse current circuit has tripped (reached the reverse current trip threshold), the current is clamped at this IROCP level. Once ROCP is activated, N-MOSFET will be turned off after 5ms. And also FLG pin pulls down after a de-glitch time of 4ms. Recovery from ROCP is automatic when the fault is removed. FLG pin pulls higher after 5ms de-glitch time.

Programming the Current-limit Threshold

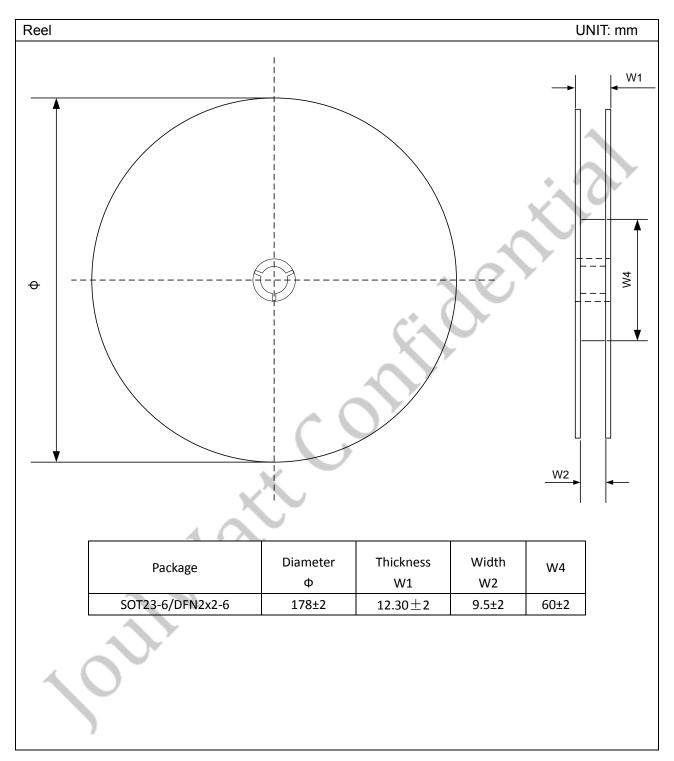
The current limit can be set by connecting resistor from the current limit adjustable pin ILIM to ground. The required value of resistor R_{LIM} for current limit is calculated as follows:

$$I_{\text{LIM}}(\text{mA}) = \frac{23950\text{V}}{\text{R}_{\text{LIM}}^{0.977}\text{K}\Omega}$$

400 65.9 66.5 65.8 67.2 333.7 396.7 500 52.5 52.3 51.8 52.8 428.2 501.6 600 43.5 43.2 42.8 43.6 520.7 604.6 700 37.2 37.4 37.0 37.8 602.4 696.0 800 32.4 32.4 32.1 32.7 698.2 800.8 900 28.7 28.7 28.4 29.0 789.0 901.5 1000 25.8 26.1 25.8 26.4 868.3 989.1 1100 23.4 23.2 23.0 23.4 981.9 1109.7 1200 21.4 21.5 21.3 21.7 1060.4 1195.4							
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Limit(mA) Low(kΩ) High(kΩ) Min(mA) (mA) M 400 65.9 66.5 65.8 67.2 333.7 396.7 1 500 52.5 52.3 51.8 52.8 428.2 501.6 1 600 43.5 43.2 42.8 43.6 520.7 604.6 1 700 37.2 37.4 37.0 37.8 602.4 696.0 1 800 32.4 32.4 32.1 32.7 698.2 800.8 1 1000 28.7 28.7 28.4 29.0 789.0 901.5 1 1100 23.4 23.2 23.0 23.4 989.1 1 1 1200 21.4 21.5 21.3 21.7 1060.4 1195.4 1	ILIM						
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600 43.5 43.2 42.8 43.6 520.7 604.6 700 37.2 37.4 37.0 37.8 602.4 696.0 800 32.4 32.4 32.1 32.7 698.2 800.8 900 28.7 28.7 28.4 29.0 789.0 901.5 1000 25.8 26.1 25.8 26.4 868.3 989.1 1100 23.4 23.2 23.0 23.4 981.9 1109.7 1200 21.4 21.5 21.3 21.7 1060.4 1195.4	475.9						
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1100 23.4 23.2 23.0 23.4 981.9 1109.7 1200 21.4 21.5 21.3 21.7 1060.4 1195.4	015.6						
1200 21.4 21.5 21.3 21.7 1060.4 1195.4	111.9						
	239.0						
1300 19.7 19.6 19.4 19.8 1164.5 1308.5	331.9						
	454.3						
1400 18.3 18.2 18.0 18.4 1251.5 1406.7	565.0						
1500 17.0 16.9 16.7 17.1 1345.5 1512.4	682.8						
1600 16.0 15.8 15.6 16.0 1436.8 1615.2	797.6						
1700 15.0 14.9 15.2 1526.5 1699.3	864.4						

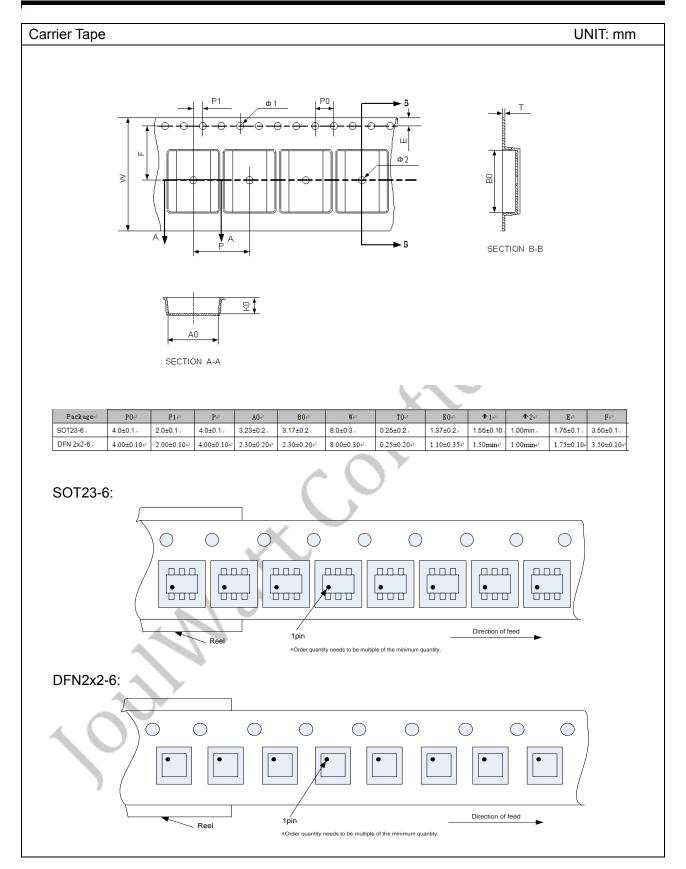
Table1. Common RLIM Resistor Selections

TAPE AND REEL INFORMATION

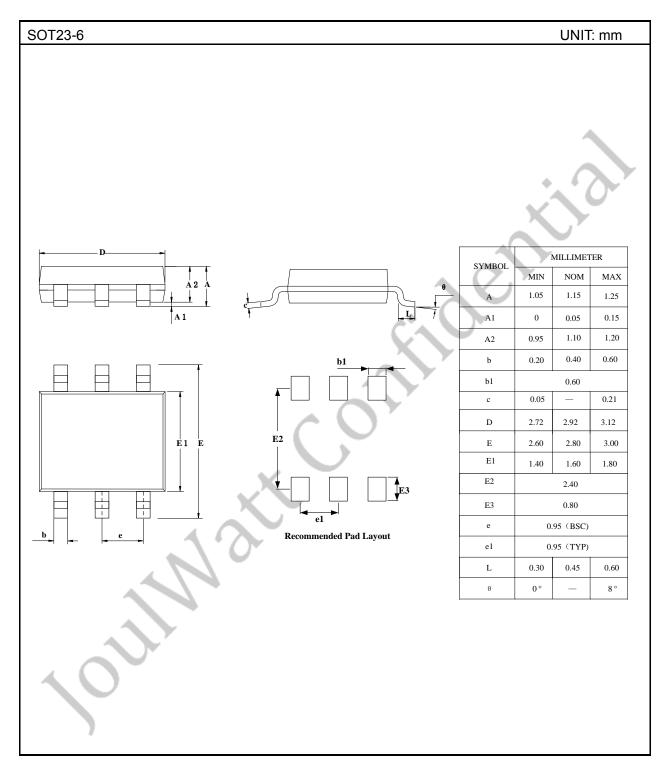


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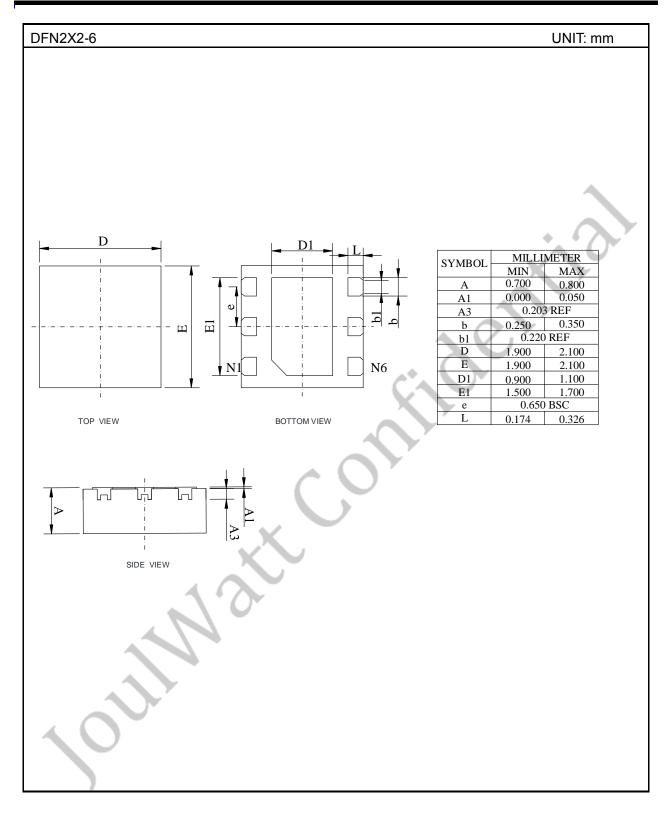
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PACKAGE OUTLINE



JWH7103



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