

## **30V 3.5A Synchronous Buck Converter**

#### **General Description**

The LP6493 is a synchronous step down regulator with CC control from a high voltage input supply. Operating with an input voltage 8V~30V, the LP6493 achieves 3.5A continuous output current with excellent load and line regulation. Current mode operation provides fast transient response and eases loop stabilization.

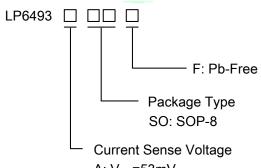
The LP6493 requires a minimum number of readily available standard external components. Other features include cable compensation, programmable current limit and thermal shutdown.

The LP6493 converters are available in the industry standard SOP8 packages.

#### **Features**

- Wide 8V to 30V Continuous Operating Input Range
- $58m\Omega/45m\Omega$  Low  $R_{DS(ON)}$  Internal Power MOSFET
- Up to 93% Efficiency
- Default 180kHz Switching Frequency
- Internal Soft Start
- Output Line Drop Compensation
- Over-Current Protection (OCP) programmable with External Resistor
- No Loop Compensation Required
- Thermal Shutdown
- Available in SOP-8 Package

## **Order Information**



A: V<sub>CS</sub>=53mV B: V<sub>CS</sub>=65mV

# LOWPOWERS EMApplications

- Car Charger / Adaptor
- Pre-Regulator for Linear Regulators
- Distributed Power Systems
- **Battery Charger**

## **Marking Information**

Device	Marking	Package	Shipping	
LP6493ASOF	LPS			
	LP6493	SOP8	4K/REEL	
LP6493BSOF	XXX			

Marking indication:

Y:Production year W:Production week X:Production batch

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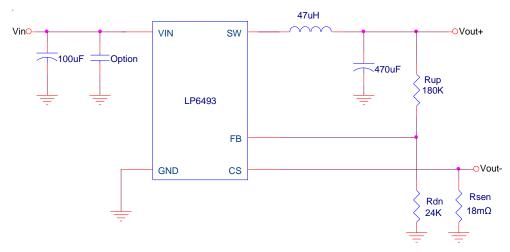
## **Functional Pin Description**

Package Type	Pin Configurations			
LP6493	FB 1 8 CS 2 7 GND 3 6 VIN 4 5	GND GND SW SW		

#### **Pin Description**

Pin	Name		Description			
1	FB		Voltage Feedback. The feedback voltage is 0.6V.			
2	CS		Output Current Sense PIN.			
3	GND		Ground.			
4	VIN		Input Supply Voltage.			
5,6	SW	LP Sen	Switch PIN.			
7,8	GND		Power Ground.			
	LowPowerSemi 微源半導體					

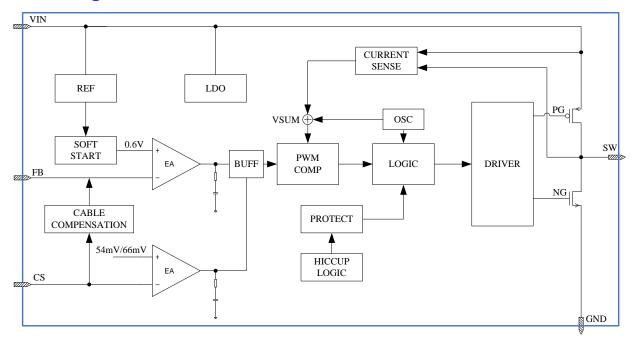
# **Typical Application Circuit**



LP6493 Application Circuit

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#### **Function Diagram**



## **Absolute Maximum Ratings Note 1**

$\diamond$	VIN to GND0.3V to +36
<b></b>	SW to GND
<b></b>	All other pin to GND
<b></b>	Storage temperature range
<b></b>	Maximum Operating Junction temperature+150°
$\Rightarrow$	+260° المحتود ا

**Note 1.** Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### **Thermal Information**

$\diamondsuit$	Maximum Power Dissipation (P <sub>D</sub> ,T <sub>A</sub> =25°C)		1.5W
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♦ Thermal Resistance (θ<sub>JA</sub>) ------ 80°C/W

## **ESD Susceptibility**

♦ HBM(Human Body Mode) ------ 2KV

♦ MM(Machine Mode) ------ 200V

## **Recommended Operating Conditions**

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#### **Electrical Characteristics**

#### V<sub>IN</sub>=12V , T<sub>A</sub>=25°C, unless otherwise noted

Parameter	Symbol	Condition	Min	Тур	Max	Units
Committee Comment	ΙQ	No switching, V <sub>FB</sub> =0.8V		0.73	1	mA
Supply Current	Isw	Switching, I <sub>load</sub> =0A		10	20	mA
VIN Under Voltage Lockout Threshold Rising	V <sub>IN_UVLO</sub>	V <sub>IN</sub> rising	6.6	7.3	8	V
VIN Under Voltage Lockout Threshold Hysteresis	V <sub>IN_HYS</sub>			1		V
Oscillator Frequency	Fsw	V <sub>FB</sub> =0.6V		180		KHz
Maximum Duty Cycle	D <sub>MAX</sub>				98	%
FB Voltage Reference	V <sub>REF</sub>		0.588	0.6	0.612	V
OVP Voltage Reference	V <sub>OVP</sub>			0.706		V
SCP voltage Reference	V <sub>REF_SCP</sub>			0.395		V
OO Vallana D. A. D. Sami		LP6493ASOF	48	53	60	mV
CS Voltage Reference	V <sub>CS</sub>	LP6493BSOF	58	65	72	mV
Line Compensation Current	I <sub>FB</sub>	erSemi 微源=	片導	1.5	2	uA
SCP Frequency	Fsw <sub>1</sub>			61		KHz
Current Limit	Ішміт			4.7		Α
High-Side Switch on Resistance	R <sub>DSON-H</sub>			58		mΩ
Low-Side Switch on Resistance	RDSON-L			45		mΩ
Thermal Shutdown	Тотр	Temp rising		150		°C
Thermal Shutdown Hysteresis	T <sub>OTP_HYS</sub>			15		°C
Hiccup Time	thiccup			260		ms
Soft-start Time	tss			2.7		ms

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## **Application Information**

The LP6493 operates by a constant frequency, current mode architecture. The output voltage is set by an external divider returned to the FB pin. An error amplifier compares the divided output voltage with a reference voltage of 0.6V and adjusts the peak inductor current accordingly.

#### **Thermal Protection**

The total power dissipation in LP6493 is limited by a thermal protection circuit. When the device temperature rises to approximately +150 °C, this circuit turns off the output, allowing the IC to cool. The thermal protection circuit can protect the device from being damaged by overheating in the event of fault conditions. Continuously running the LP6493 into thermal shutdown degrades device reliability.

#### **Current Limit**

The Current limit is set by outside resistance (RSEN), When the CS voltage larger than 53mV/65mV, the current limit is happened that driver can be turned off. The current limit set according to the following equation:

 $I_{OUT} = V_{CS} / R_{SEN}$ 

#### **Setting Output Voltage**

The output voltage is set with a resistor divider from the output node to the FB pin. It is recommended to use divider resistors with 1% tolerance or better. To improve efficiency at very light loads consider using larger value resistors. If the values are too high the regulator is more susceptible to noise and voltage errors from the FB input current are noticeable. For most applications, a resistor in the  $10k\Omega$  to  $1M\Omega$ range is suggested for R<sub>UP</sub> and R<sub>DN</sub>. The output voltage is established by the following equations

$$V_{OUT} = 0.6V \cdot (1 + R_{UP} / R_{DN})$$

Where V<sub>FB</sub> is 0.6V.

#### **Output Cable Resistance Compensation**

To compensate for resistive voltage drop across the charger's output cable, the LP6493 integrates a simple, user-programmable cable voltage drop compensation using the impedance at the FB pin. The delta V<sub>OUT</sub> rises when the feedback resistance R<sub>SEN</sub> value rises, use the equation below:

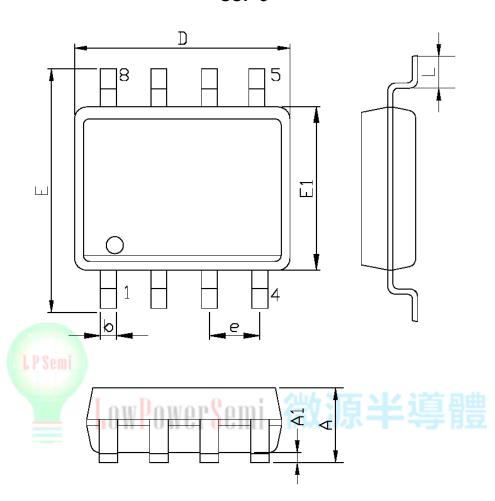
 $\Delta V_{OUT}(mV)$ = $I_{OUT}(A)$ • $R_{SEN}(m\Omega)$ • $R_{UP}(K\Omega)$  /40  $K\Omega$ 

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# **Packaging Information**





SYMBOLS	MILLIM	ETERS	INCHES		
	MIN.	MAX.	MIN.	MAX.	
A	1.35	1.75	0.053	0.069	
A1	0.10	0.25	0.004	0.010	
D	4.9	90	0.193		
Е	5.80	6.20	0.228	0.244	
E1	3.9	90	0.1	53	
L	0.40	1.27	0.016	0.050	
ь	0.31	0.51	0.012	0.020	
e	1.3	27	0.050		

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