

**Features**

- Excellent Broadband Mixer Driver
- Single Ended Fed Doubler with Distributed Buffer Amplifier
- Excellent LO Driver for MACOM Receivers
- +15 dBm Output Drive
- 100% On-Wafer RF, DC and Output Power Testing
- 100% Visual Inspection to MIL-STD-883 Method 2010
- RoHS\* Compliant

**Applications**

- Point-to-Point Radio
- Microwave
- LMDS
- SATCOM
- VSAT

**Description**

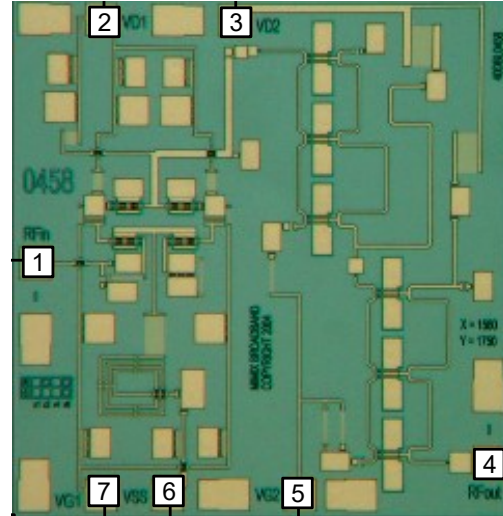
This single ended fed (no external balun required) 7.5 - 25.0 / 15.0 - 50.0 GHz GaAs MMIC doubler has a 15 dBm output drive and is an excellent LO doubler that can be used to drive fundamental mixer devices. It is also well suited to drive MACOMs' XR1002 receiver device.

This MMIC uses a GaAs pHEMT device model technology, and is based upon electron beam lithography to ensure high repeatability and uniformity. The chip has surface passivation to protect and provide a rugged part with backside via holes and gold metallization to allow either a conductive epoxy or eutectic solder die attach process.

**Ordering Information**

Part Number	Package
XX1000-BD-000V	vacuum release gel paks
XX1000-BD-EV1	evaluation board

**Chip Device Layout**



**Pad Configuration<sup>1</sup>**

Pad	Function	Description
1	RF <sub>IN</sub>	RF Input
2	V <sub>D1</sub>	Drain Voltage Stage 1
3	V <sub>D2</sub>	Drain Voltage Stage 2
4	RF <sub>OUT</sub>	RF Output
5	V <sub>G2</sub>	Gate Voltage Stage 2
6	V <sub>SS</sub>	Source Supply Voltage
7	V <sub>G1</sub>	Gate Voltage Stage 1

1. Backside metal is RF, DC and thermal ground.

\* Restrictions on Hazardous Substances, compliant to current RoHS EU directive.

# Active Doubler

## 7.5 - 25.0 / 15.0 - 50.0 GHz



XX1000-BD  
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### Electrical Specifications: Input Freq. = 7.5 - 25 GHz, $T_A = 25^\circ\text{C}$

Parameter	Units	Min.	Typ.	Max.
Output Frequency Range	GHz	15	—	50
Input Return Loss	dB	—	12	—
Output Return Loss	dB	—	12	—
Harmonic Gain	dB	—	13	—
Fundamental Rejection	dBc	—	20	—
Saturated Output Power	dBm	—	15	—
RF Input Power	dBm	-10	—	+10
Output Power at 0 dBm $P_{IN}$	dBm	—	13	—
Drain Bias Voltage ( $V_{D1,2}$ )	VDC	—	5.0	5.5
Gate Bias Voltage ( $V_{G1}$ )	VDC	-1.2	-0.6	+0.1
Gate Bias Voltage ( $V_{G2}$ )	VDC	-1.2	0.0	+0.1
Drain Current ( $I_{D1,2}$ ) ( $V_D = 5.0\text{ V}$ , $V_{G1} = -0.6\text{ V}$ , $V_G = 0\text{ V}$ Typical)	mA	—	265	280
Source Voltage ( $V_{SS}$ )	VDC	-5.5	-5.0	-2.0
Source Current ( $I_{SS}$ )	mA	25	50	60

### Absolute Maximum Ratings<sup>2</sup>

Parameter	Absolute Maximum
Drain Voltage ( $V_{D1}$ , $V_{D2}$ )	+6 V
Source Voltage ( $V_{SS}$ )	-6 V
Drain Current ( $I_{D1}+I_{D2}$ )	320 mA
Source Current ( $I_{SS}$ )	60 mA
Gate Bias Voltage ( $V_{G1}$ )	+0.3 V
Gate Bias Voltage ( $V_{G2}$ )	+0.1 V
RF Input Power	+12 dBm
Storage Temperature	-65°C to +165°C
Operating Temperature	-55°C to MTTF Table
Channel Temperature	MTTF Table

2. Channel temperature directly affects a device's MTTF. Channel temperature should be kept as low as possible to maximize lifetime.

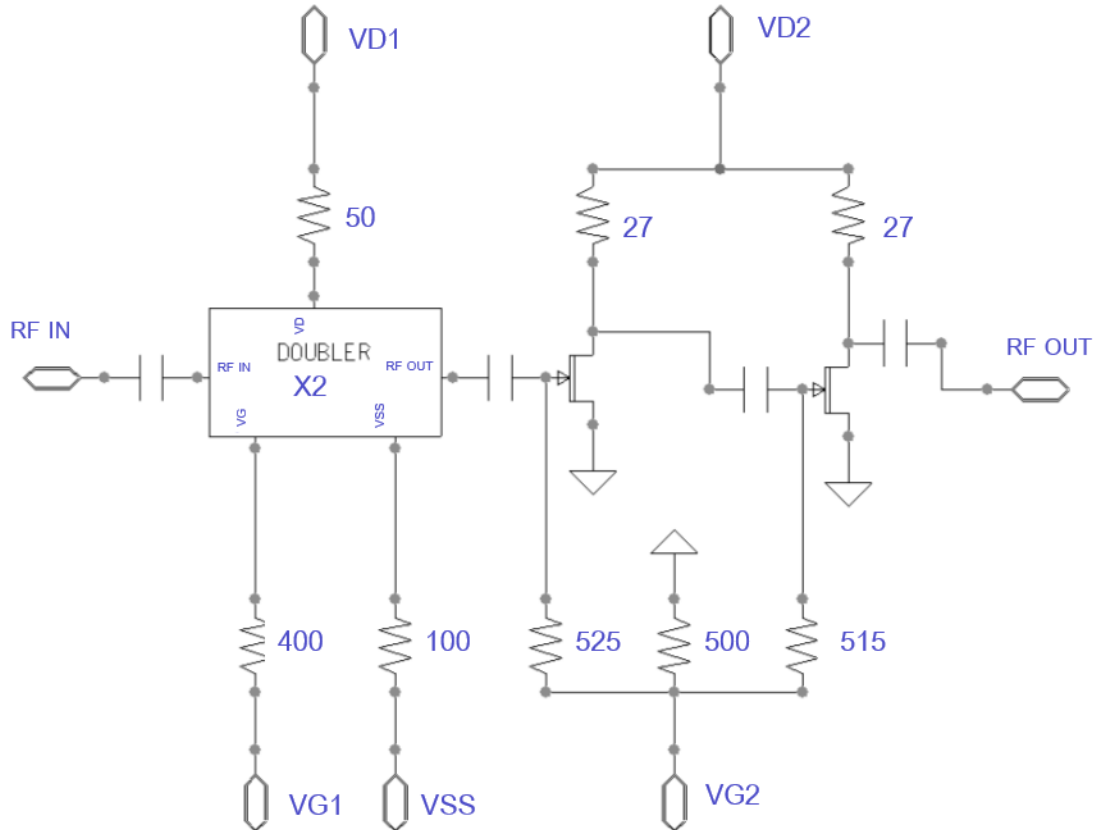
### Handling Procedures

Please observe the following precautions to avoid damage:

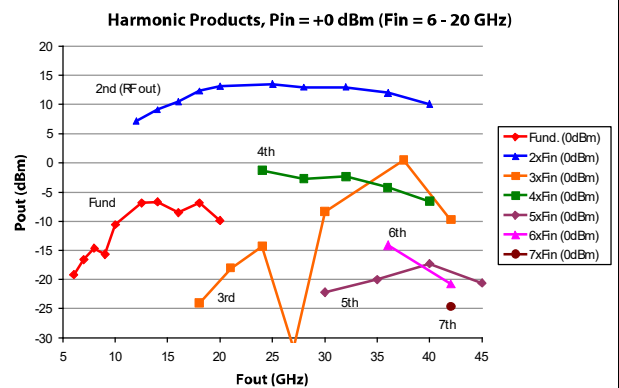
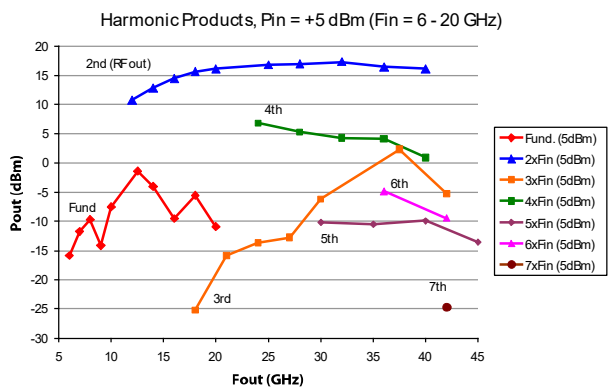
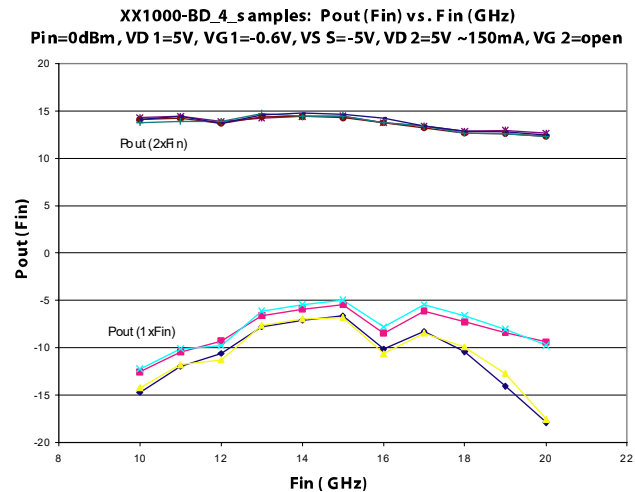
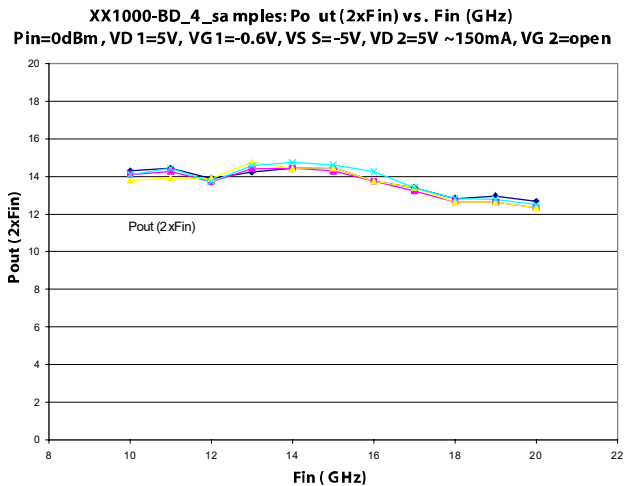
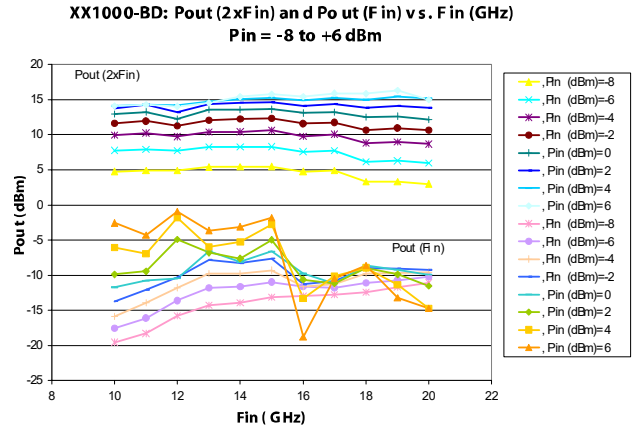
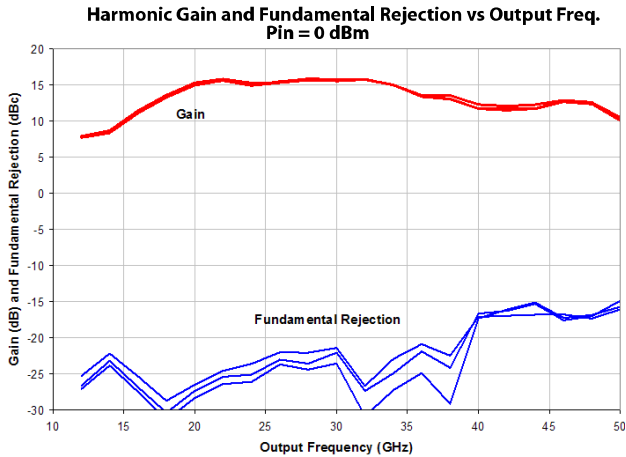
### Static Sensitivity

These electronic devices are sensitive to electrostatic discharge (ESD) and can be damaged by static electricity. Proper ESD control techniques should be used when handling these Class 2 devices.

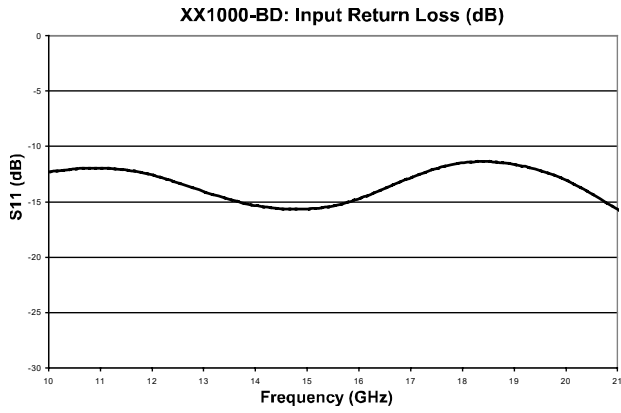
Block Diagram & Schematics



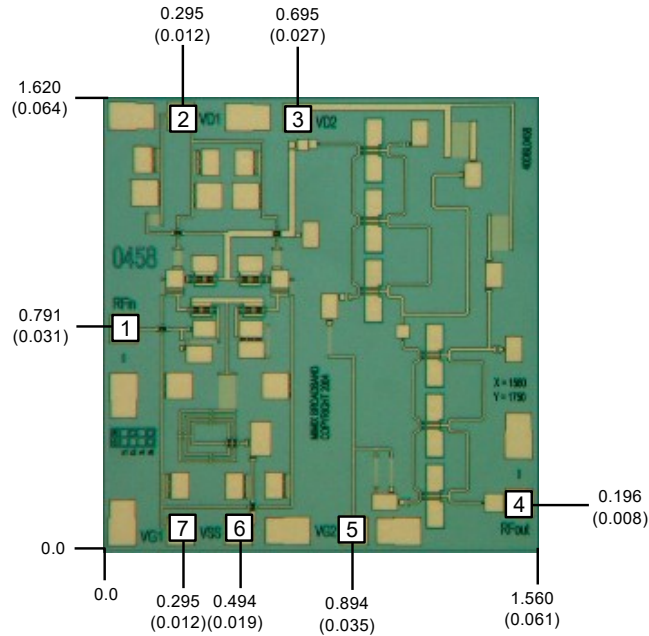
### Typical Performance Curves



Typical Performance Curves (cont.)



### Mechanical Drawing

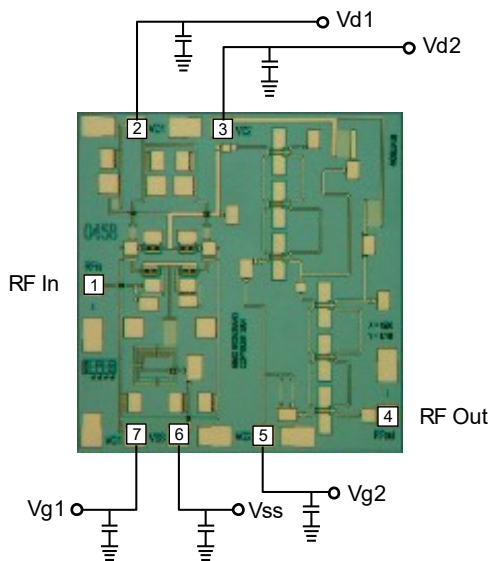


(Note: Engineering designator is 40DBL0458)

Units: millimeters (inches) Bond pad dimensions are shown to center of bond pad.  
 Thickness: 0.110 +/- 0.010 (0.0043 +/- 0.0004), Backside is ground, Bond Pad/Backside Metallization: Gold  
 All Bond Pads are 0.100 x 0.100 (0.004 x 0.004).  
 Bond pad centers are approximately 0.109 (0.004) from the edge of the chip.  
 Dicing tolerance: +/- 0.005 (+/- 0.0002). Approximate weight: 1.566 mg.

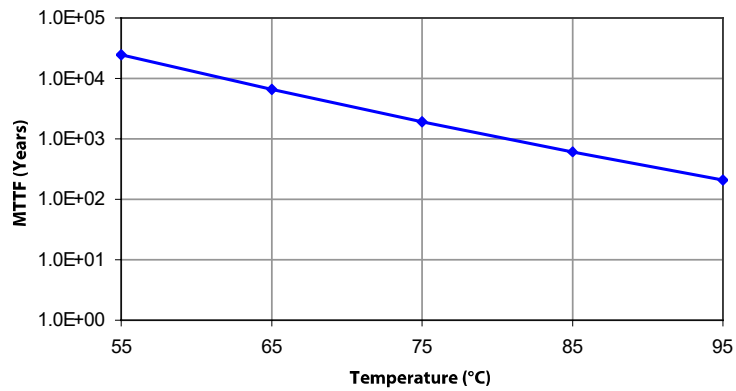
Bond Pad #1 (RF In)      Bond Pad #3 (Vd2)      Bond Pad #5 (Vg2)      Bond Pad #7 (Vg1)  
 Bond Pad #2 (Vd1)      Bond Pad #4 (RF Out)      Bond Pad #6 (Vss)

### Bias Arrangement



6 **Bypass Capacitors** - See App Note [2]

### MTTF vs. Back-plate Temperature (°C)



MTTF is calculated from accelerated life-time data of single devices and assumes isothermal back-plate.

Bias Conditions:  $V_{D1,2} = 5\text{ V}$ ,  $I_{D1,2} = 220\text{ mA}$ ,  $V_{SS} = -5\text{ V}$ ,  $I_{SS} = 50\text{ mA}$

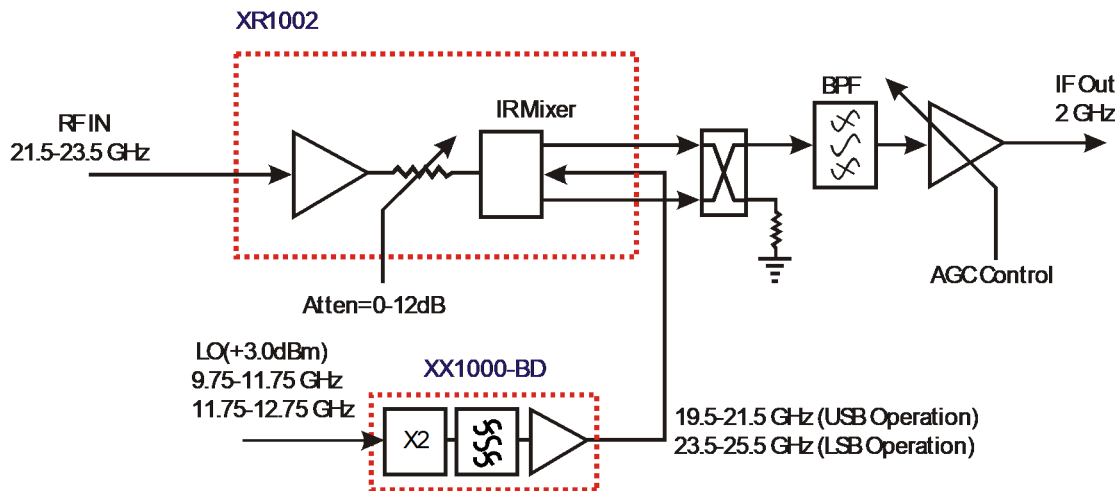
### App Note [1] Biasing -

It is recommended to separately bias each doubler stage with fixed voltages of  $V_{D1,2} = 5\text{ V}$ ,  $V_{SS} = -5\text{ V}$  and  $V_{G1} = -0.6\text{ V}$ . The typical DC currents are  $I_{D1} = 80\text{ mA}$ ,  $I_{D2} = 140\text{ mA}$  and  $I_{SS} = 50\text{ mA}$ .  $V_{G2}$  can be used for active control biasing of  $V_{D2}$ , or it can be left open and  $V_{D2}$  will self bias at approximately 140 mA. Maximum output power is achieved with  $V_{SS} = -5\text{ V}$  and  $I_{SS} = 50\text{ mA}$  but the device will operate with reduced bias to  $V_{SS} = -2\text{ V}$  and  $I_{SS} = 25\text{ mA}$ . It is also recommended to use active biasing on  $V_{D2}$  with  $V_{G2}$  to keep the currents constant as the RF power and temperature vary; this gives the most reproducible results. Depending on the supply voltage available and the power dissipation constraints, the bias circuit may be a single transistor or a low power operational amplifier, with a low value resistor in series with the drain supply used to sense the current. The gate of the pHEMT is controlled to maintain correct drain current and thus drain voltage. The typical gate voltage for  $V_{G2} = -0.1\text{ V}$ . Typically the gate is protected with silicon diodes to limit the applied voltage. Also, make sure to sequence the applied voltage to ensure negative gate bias is available before applying the positive drain supply.

### App Note [2] Bias Arrangement -

For individual stage bias (recommended for doubler applications) - Each DC pad ( $V_{D1,2}$ ,  $V_{SS}$  and  $V_{G1,2}$ ) needs to have DC bypass capacitance ( $\sim 100 - 200\text{ pF}$ ) as close to the device as possible. Additional DC bypass capacitance ( $\sim 0.01\text{ }\mu\text{F}$ ) is also recommended.

## Typical Application



MMIC based 18 - 34 GHz Double / Receiver Block Diagram  
(changing LO and IF frequencies as required allows the design to operate as high as 34 GHz.)



# Active Doubler

## 7.5 - 25.0 / 15.0 - 50.0 GHz



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Rev. V3

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