Evaluates: MAX20047

General Description

The MAX20047 evaluation kit (EV kit) demonstrates the MAX20047 IC in an integrated and small package format.

The EV kit is configured for 400kHz operation, with a 3A (min) per-port current limit. There are two Type A receptacles that can be used to demonstrate the charger emulation capabilities.

Features

- Configurable Charge-Detection Modes
 - USB-IF BC 1.2 Dedicated Charging Port (DCP)
 - Apple® 2.4A, 2.1A, 1.0A, and Samsung® Termination Resistors
- Integrated, High-Efficiency, DC-DC Converter (440kHz to 2.2MHz)
- Proven PCB Layout
- Fully Assembled and Tested

Quick Start

Test Points

Figure 1 depicts the connections and test points available.



Figure 1. MAX20047 Evaluation Kit Test Points

Ordering Information appears at end of data sheet.

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Samsung is a registered trademark of Samsung Electronics Co., Ltd.



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Detailed Description

The MAX20047 EV kit comes fully assembled, tested, and installed with a MAX20047AFPA/V+ IC, which includes a dual-package external FET added for short-to-battery protection.

EV Kit Interface

HVEN Terminal

The HVEN terminal connects to the IC's HVEN pin through header J3. The HVEN pin is compatible with a wide range of input voltages, from +2.4V (logic-level high) to +40V (automotive battery level).

Header/Jumper (J3)

The J3 header/jumper selects the input source for the HVEN pin. Installing a jumper connects the HVEN pin directly to the VBAT terminal. With this selection, applying a voltage \geq 5.8V enables the IC. **Note:** There is a 2s delay before the BUS voltage appears at the HVBUS1 and HVBUS2 terminals. See <u>Table 1</u> for J3 jumper selections.

Alternatively, removing the jumper connects the HVEN pin to the HVEN test point. The HVEN test point is lightly pulled to ground with a $100k\Omega$ resistor (R5). With this selection, the IC is disabled. With the jumper on J3 removed, a

logic-level signal $\ge 2.4V$ connected to the HVEN test point enables the IC. A source voltage $\ge 5.8V$ must be applied to the VBAT terminal. This configuration permits an MCU GPIO signal to control the IC's HVEN pin.

VBAT/PGND Terminals

Connect the battery voltage input between VBAT and PGND. The IC's DC-DC converter output voltage can be measured on the VOUT test point (see Figure 1).

HVBUS1/HVBUS2 Terminals

The BUS output voltage for each USB port can be measured on the test points HVBUS1/HVBUS2, respectively.

Basic Evaluation Procedures

Testing HVBUS1/HVBUS2 Voltage

Figure 2 shows the test setup to evaluate the IC's VBUS port voltages:

Table 1. Jumper Selections (J3)

REFERENCE DESIGNATOR	INSTALLED	REMOVED
J3	HVEN pin connected to VBAT	HVEN pin pulled to PGND through a 100kΩ resistor (R5)



Figure 2. Test Setup for Testing HVBUS1/HVBUS2

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Equipment Required for Test

- One bench supply (12V/3A)
- Two multimeters (M1, M2)

Procedure

- 1) With the EV kit connected (see Figure 2), set PS1 to 12.000V and enable the power-supply output.
- 2) Wait at least 3s for the IC's port control logic to validate a successful startup condition
- 3) Observe that multimeters M1 and M2 indicate \approx 5.2V.

Testing DCP Charger Detection (iPhone[®])

Figure 3 shows the test setup to evaluate the IC's VBUS port voltages.

Equipment Required for Test

- Bench supply (12V/3A)
- In-line USB voltage/current meter (DROK or similar)
- iPhone

Procedure

- 1) Connect the equipment (see Figure 3), set PS1 to 12.000V, and enable the power supply output.
- Wait at least 3s for the IC's port control logic to validate a successful startup condition. Observe that M1 and M2 indicate ≈ 5.2V.



Figure 3. Test Setup for Verifying DCP Charger Detection with an iPhone

iPhone is a registered trademark of Apple Inc.

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Testing DCP Charger Detection (iPad[®])

Figure 4 shows the test setup to evaluate the IC's VBUS port voltages:

Equipment Required for Test

- Bench supply (12V/3A),
- In-line USB voltage/current meter (DROK or similar), and an iPad.

Procedure

- 1) Connect the equipment (see Figure 4), set PS1 to 12.000V, and enable the power-supply output.
- Wait at least 3s for the IC's port control logic to validate a successful startup condition. Observe that M1 and M2 indicate ≈ 5.2V.



Figure 4. Test Setup for Verifying DCP Charger Detection with an iPad

iPad is a registered trademark of Apple Inc.

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Advanced Evaluation Procedures

Testing the VBUS Current Limit

The VBUS current limit for both USB ports are factory configured for 3.3A, which can be changed to 2.75A, 2.41A, or 2.1A by selecting a different R_{ILIM} resistor. See the <u>Other Configurations</u> section for details on making this selection. Figure 5 shows the test setup to evaluate the IC's VBUS current limit:

Equipment Required

- Bench supply (12V/3A)
- Two multimeters (current measurement up to 5A)
- Electronic load (up to 5A)

Procedure

1) Connect the equipment (see Figure 5), set PS1 to

12.000V, and enable the power-supply output.

- Wait at least 3s for the IC's port control logic to complete a successful startup. Observe that M2 indicates ≈ 5.2V.
- Slowly ramp the electronic load sink current from 0.0A to 3.5A. Observe the current indicated on M1 and voltage on M2.
- 4) At some point between the 3.1A and 3.3A sink current, the current-limit threshold is exceeded, and the IC's VBUS voltage to the port is shut down. The voltage indicated on M2 should fall to zero, and the current flow indicated on M1 should also fall to zero.
- 5) Reduce the electronic load sink current below 3.0A.
- 6) After 3s the VBUS voltage should return, and the port again sources current to the electronic load.



Figure 5. Test Setup for Testing DCP Current Limit

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Testing Short-to-Ground Fault

The IC independently protects the HVBUS1/HVBUS2 channels against shorts to ground. Figure 6 shows the test setup to evaluate VBUS short-to-ground fault protection.

Equipment Required

- Bench supply (12V/3A)
- Multimeter
- Short jumper wire (18AWG)

Procedure

1) Connect the equipment (see Figure 6), set PS1 to 12.000V, and enable the power-supply output.

- Wait at least 3s for the IC's port control logic to complete a successful startup. Observe that M1 indicates ≈ 5.2V.
- 3) Connect a jumper wire between the PGND and HVBUS2 terminals.
- Observe that HVBUS1 voltage falls to zero, indicating that the IC's short-to-ground protection has been triggered. After ≈ approximately 3s, the 5.2V on the HVBUS1 terminal should return.
- 5) Remove the short between HVBUS2 and PGND, observing no change in the HVBUS1 voltage.
- 6) After ≈ 3s, the 5.2V on the HVBUS2 terminal should return.



Figure 6. Test Setup for Verifying VBUS Short-to-Ground Protection

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Testing Short-to-Battery Fault

The IC independently protects the HVBUS1 and HVBUS2 channels against shorts to battery. <u>Figure 7</u> shows the test setup to evaluate VBUS short-to-battery fault protection.

Equipment Required

- Bench supply (12V/3A)
- One multimeter (M1)
- Short jumper wire (18AWG)

Procedure

1) Connect the equipment (see Figure 7), set PS1 to 12.000V, and enable the power-supply output.

- Wait at least 3s for the IC's port control logic to complete a successful startup. Observe that M1 indicates ≈ 5.2V.
- 3) Connect a jumper wire between the VBAT and HVBUS2 terminals.
- 4) Observe that HVBUS1 voltage falls to zero, indicating that the IC's short-to-battery protection has been triggered.
- 5) Remove the shorting jumper wire. Observe M1. When the overvoltage charge on the HVBUS1 output capacitor decays enough, the IC's protection logic initiates a recovery cycle, and after ≈ 5.200V, returns to the HVBUS1 and HVBUS2 terminals.



Figure 7. Test Setup for Verifying VBUS Short-to-Battery Protection

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Other Configurations

FOSC Resistor Selection (RFOSC)

The switching frequency of the IC's DC-DC is set by connecting a resistor between the FOSC pin and AGND. The internal oscillator can be tuned across a wide frequency range, providing greater system design flexibility. The graph shown in Figure 8 plots frequency vs. R_{FOSC} resistance. See Table 2 for resistor values for selected switching frequencies.

CONFIG Resistor Selection (RCONFIG)

During device boot, the hold configuration is loaded with a value depends on decoding/reading the external CONFIG resistor. For more details regarding the value loaded as a function of resistors connected (see Table 3).

Setting the VBUS Current Limit (RILIM)

ILIM is a multi-functional pin that can be used to program the VBUS current limit, Apple divider-current, and foldback-current threshold. <u>Table 4</u> lists configuration options for the ILIM pin.



Figure 8. Frequency vs. RFOSC Resistance

Table 2. RFOSC Resistor Values for Selected Switching Frequency

FOSC (kHz)	224,320	300,840	380,500	493,100	595,800	756,000	1,000,500	1,479,000	2,237,000
R _{FOSC} (kΩ)	133	97.6	73.2	59	48.7	38.3	28.7	19.1	12.7

Table 3. Configuration Pin Options

CONFIGURATION PIN CONNECTION	LEVEL (at 50µA)	HOLD MODE	HOLD TIME (MINUTES)
Connect to AGND	Ground	Disabled	N/A
24,900Ω to AGND	1.25V	Enabled	30
Connect to BIAS	VBIAS	Enabled	60

Table 4. ILIM Pin Options

ILIM PIN CONNECTION	APPLE DIVIDER-CURRENT (A)	ILIMIT (A)	FOLDBACK (A)
Connect to AGND	2.4	3.3	None
8,870Ω to AGND	2.4	3.3	2.41
15,800Ω to AGND	2.4	3.3	2.1
24,900Ω to AGND	2.4	2.75	2.1
35,700Ω to AGND	2.4	2.75	None
49,900Ω to AGND	2.1	2.41	None
68,100Ω to AGND	2.1	2.41	2.1
Connect to BIAS	1.0	2.1	None

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PCB Layout Guidelines

Good PCB layout is critical to proper system performance. The loop area of the DC-DC conversion circuitry must be minimized as much as possible. Place the input capacitor, power inductor, and output capacitor very close to the IC. Shorter traces should be prioritized over wider traces. Similarly, the COMP network should be close to the IC and connect directly to AGND. The BIAS capacitor should be close to the IC and connect directly to PGND or via down to a layer 2 ground plane.

Alow-impedance ground connection between the input and output capacitors is necessary (route through the ground pour). Treat the SUP and PGND pins the same as you would an exposed pad. Place multiple vias in the pad and connect to all other ground layers for proper heat dissipation; failure to perform this step can result in the IC repeatedly reaching thermal shutdown. Prioritize the use of a large, single ground as opposed to multiground schemes; high-frequency return currents flow directly under the corresponding traces. Proper thermal dissipation is critical for this application. Maximize ground pour and keep a single continuous pour on the top and bottom layers; do not isolate pours. Most heat radiates from PGND and SUP, so adding internal and bottom-layer SUP pours can be beneficial. Do not place noncritical traces and components near the IC footprint (on any layer). This allows for wide/large copper pour near the PGND and SUP vias. Contact the Maxim applications team for layout reviews and recommendations.

Ordering Information

PART	TYPE
MAX20047EVKIT#	EV Kit

#Denotes RoHS compliant.

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REFERENCE DESIGNATOR	QTY.	NAME	DESCRIPTION	MFG. PART NUMER	MANUFACTURER
C1,C2	2	C_OUT	Ceramic Capacitor (1210) 47uF 10V X7R	GRM32ER71A476KE15L	Murata
C3	1	C_IN	Electrolytic Capacitor(8x6.2mm) 47uF 20% 35V SMD	EEEFC1V470P	Panasonic
C4	1	C_SUPSW	Ceramic Capacitor (1206) 10uF 35V X7R	CGA5L1X7R1V106K160AC	TDK
C5, C6	2	C_SDMOS1/2	Ceramic Capacitor (0805) 1uF 16V X7R	GCM219R71C105KA37J	Murata
C7,C8,C12	3	C_FILT_OUT, C_BST	Ceramic Capacitor (0402) 0.1uF 50V X7R	CGA2B3X7R1H104K050BB	TDK
C9	1	C_COMP_P	Ceramic Capacitor (0402) 3pF 50V C0G	CGA2B2C0G1H3R3C050BA	TDK
C10	1	C_COMP_S - 400kHz	Ceramic Capacitor (0402) 2200pF 50V X7R	GCM155R71H222KA37D	Murata
C11	1	C_BIAS	Ceramic Capacitor (0805) 2.2uF 16V X7R 0805	CGA4J3X7R1C225K125AB	TDK
C13,C14	2	C_FILT_IN	Ceramic Capacitor (0402) 2200pF 50V X7R	GCM155R71H222KA37D	Murata
C15	1	CSUP_FILT	Ceramic Capacitor (0603) 0.1uF 50V X7R	CGA3E2X7R1H104K080AA	TDK
C16,C17	2	C_FILT_OUT	Ceramic Capacitor (0402) 0.22uF 35V X7R	CGA2B1X7R1V224KC	TDK
C18, C19	2	C_HVBUS1/2	Ceramic Capacitor (1206) 10uF 25V X7R	CGA5L1X7R1E106M160AC	TDK
FB1-3	3	Input/Output Ferrite	Ferrite Bead (1806) Automotive Grade, 6 Amp Max, 9 mOhms DCR	BLM41PG600SZ1L	Murata
GND1/2	2	Multipurpose Test Point	Black Test Point	5011	Keystone
HVBUS1/2	2	Multipurpose Test Point	Red Test Point	5010	Keystone
HVEN	1	Multipurpose Test Point	Yellow Test Point	5014	Any (Keystone)
J1,J2	2	FCI_73725_USB_Rec	Type A USB Receptacle, Through Hole, Right Angle	73725	Amphenol FCI
J3	1	HVEN_VBAT	2 Pos. 0.100" Header	4-103327-0	Any (TE)
L1	1	L - 400kHz	4.7uH Molded Inductor, 10.5A Isat	XAL6060-472MEB	Coilcraft
OUT	1	Miniature Test Point	DNP	DNP	—
PGND, VBAT	2	Test Point	Wire Loop (18 guage uncoated bus bar)		Any (Belden)
Q1	1	External FETs	Dual N-Channel Mosfet (8SOIC) 30V 4.9A	NTMD4820N	On Semi
R1	1	R_COMP_S - 400kHz	Resistor (0402) 16.2kΩ 1% 1/16W	CRCW040216K2FKED	Vishay Dale
R2	1	R_ILIM	Resistor (0402) SHORT	CRCW04020000Z0ED	Vishay Dale
R3	1	R_FOSC	Resistor (0402)73.2k 1% 1/16W	CRCW040273K2FKED	Vishay Dale
R4	1	R_CONFIG	Resistor (0402) SHORT	CRCW04020000Z0ED	Vishay Dale
R5	1	R_HVEN	Resistor (0402) 100kΩ 5% 1/16W	CRCW0402100KJNED	Vishay Dale
R6, R7	2	R_HVBUS1/2	Resistor (0603) 1Ω ±5% 1/10W	CRCW06031R00JNEA	Vishay Dale
U1	1	Maxim Device	Dual DCP Charging Port	MAX20047AFPA/V+	Maxim Integrated
_	1		PCB: MAX20047 Evaluation Kit	MAX20047EVKIT#	Maxim Integrated

MAX20047 EV Kit Bill of Materials

MAX20047 EV Kit Schematic



MAX20047 EV Kit Schematic

MAX20047 EV Kit PCB Layouts



MAX20047 EV Kit Component Placement Guide—Top Silkscreen

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MAX20047 EV Kit PCB Layout—Top Layer

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MAX20047 EV Kit PCB Layout—Layer 2

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MAX20047 EV Kit PCB Layout—Layer 3

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MAX20047 EV Kit PCB Layout—Bottom Layer

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MAX20047 EV Kit Component Placement Guide—Bottom Silkscreen

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Revision History

REVISION	REVISION	DESCRIPTION	PAGES
NUMBER	DATE		CHANGED
0	3/18	Initial release	—

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