

## MAX22203

# 65V, 3.8A Dual Brushed or Single Stepper Motor Driver with Integrated Current Sense

### General Description

The MAX22203 is a dual 65V, 3.8A<sub>MAX</sub> H-Bridge with PWM inputs and accurate Current Drive Regulation (CDR). Each H-Bridge can be controlled individually and has a very low typical R<sub>ON</sub> (high-side + low-side) of 0.3Ω, resulting in high driving efficiency and low heat generation. The MAX22203 can be used to drive two Brushed DC motors or a single Stepper Motor.

The integrated CDR can limit the start up or stall the current of a Brushed DC motor or control the phase current for stepper operation.

The bridge output current is sensed by a non-dissipative Integrated Current Sensing (ICS) eliminating the bulky external power resistors (which are normally required for this function) and compared with a configurable threshold current (I<sub>TRIP</sub>). The I<sub>TRIP</sub> threshold can be set independently for the two full bridges by connecting the external resistors to pins R<sub>EFA</sub> and R<sub>EFB</sub>.

The maximum output current per H-Bridge is I<sub>MAX</sub> = 3.8A and is limited by the Overcurrent Protection (OCP) circuit. This current can be driven for very short transients and is aimed to effectively drive small capacitive loads. The maximum user-configurable current regulation threshold is I<sub>TRIP\_MAX</sub> = 3A. The maximum RMS current (I<sub>RMS</sub>) per H-Bridge is 2A<sub>RMS</sub> on a standard JEDEC 4-layer board. The maximum RMS current can be limited by thermal considerations and depends on the thermal characteristic of the application (PCB ground planes, heat sinks, forced air ventilation, etc).

The MAX22203 features Overcurrent Protection (OCP), Thermal Shutdown (TSD), and Undervoltage Lockout (UVLO). An open-drain active low nFAULT pin is activated every time a fault condition is detected. During Thermal Shutdown and Undervoltage Lockout, the driver is tristated until normal operations are restored.

The MAX22203 is packaged into a small TQFN38 5mm x 7mm package.

### Applications

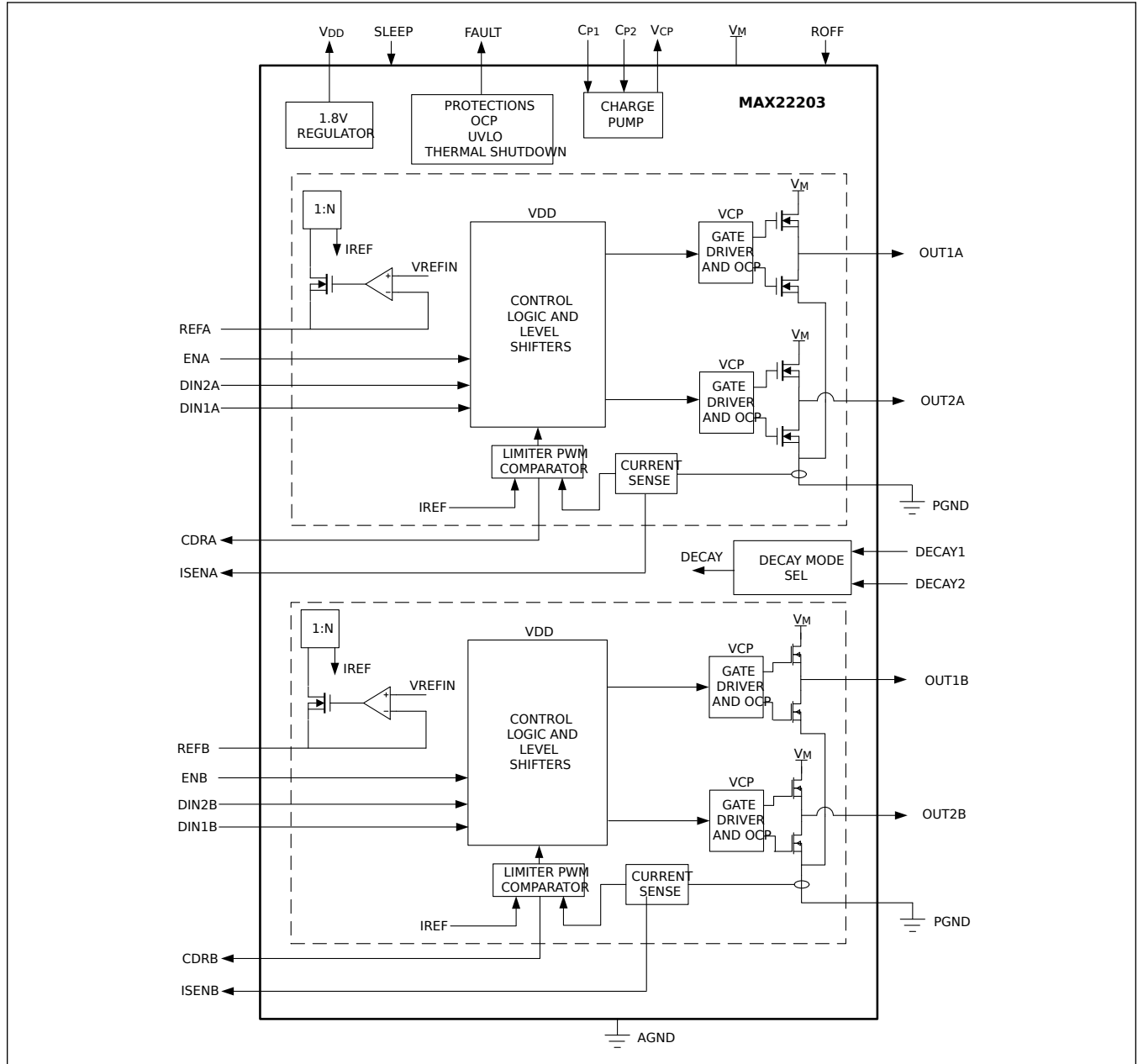
- Brushed DC Motor Driver
- Stepper Motor Driver
- Solenoid Driver
- Latched Valves

### Benefits and Features

- Two H-Bridges with 65V Maximum Operating Voltage
  - Total R<sub>ON</sub> (High-Side + Low-Side): 300mΩ typical (T<sub>A</sub> = 25°C)
- Current Ratings Per H-Bridge (Typical at 25°C):
  - I<sub>MAX</sub> = 3.8A (Impulsive Current for Driving Capacitive Loads)
  - I<sub>TRIP\_MAX</sub> = 3A (Maximum Current Setting for Internal Current Drive Regulation)
  - I<sub>RMS</sub> = 2A<sub>RMS</sub>
- Integrated Current Drive Regulation (CDR)
  - Internal Current Sensing (ICS) Eliminates External Bulky Resistors and Improves Efficiency
  - Current Drive Regulation Monitor Output Pins (CDRA and CDRB)
  - Multiple Decay Modes (Slow, Mixed, Fast)
  - Fixed Off Time Configurable with External Resistance.
- Current-Sense Output (Current Monitor)
- Fault Indicator Pin (FAULT)
- Protections
  - Overcurrent Protection for Each Individual Channel (OCP)
  - Undervoltage Lockout (UVLO)
  - Thermal Shutdown (TSD) T<sub>J</sub> = +165°C
- TQFN38 5mm x 7mm package (TSSOP38 4.4mm x 9.7mm available in the future)

**Ordering Information appears at end of data sheet.**

Simplified Block Diagram



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MAX22203

65V, 3.8A Dual Brushed or Single Stepper Motor  
Driver with Integrated Current Sense

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## Absolute Maximum Ratings

$V_M$ to GND .....	-0.3V to +70V	ROFF to GND .....	-0.3V to min(+2.2V, $V_{DD} + 0.3V$ )
$V_{DD}$ to GND .....	-0.3V to min(+2.2V, $V_M + 0.3V$ )	ISEN_ to GND .....	-0.3 to min(+2.2V, $V_{DD} + 0.3V$ )
PGND to GND .....	-0.3V to +0.3V	DIN_ to GND .....	-0.3V to 6V
OUT_ .....	-0.3V to $V_M + 0.3V$	EN_ to GND .....	-0.3V to 6V
$V_{CP}$ to GND .....	$V_M - 0.3V$ to min(+74V, $V_M + 6V$ )	DECAY_ to GND .....	-0.3V to 6V
CP2 to GND .....	$V_M - 0.3V$ to $V_{CP} + 0.3V$	SLEEP to GND .....	-0.3V to Min(+70V, $V_M + 0.3V$ )
CP1 to GND .....	-0.3V to $V_M + 0.3V$	Operating Temperature Range .....	-40°C to 125°C
FAULT to GND .....	-0.3V to 6V	Junction Temperature .....	+150°C
CDR_ to GND .....	-0.3V to 6V	Storage Temperature Range .....	-65°C to +150°C
REF_ to GND .....	-0.3V to min(+2.2V, $V_{DD} + 0.3V$ )	Soldering Temperature (reflow) .....	260°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## Package Information

### TQFN 38 - 5x7mm

Package Code	T3857-1C
Outline Number	<a href="#">21-0172</a>
Land Pattern Number	<a href="#">90-0076</a>
<b>Thermal Resistance, Single-Layer Board:</b>	
Junction to Ambient ( $\theta_{JA}$ )	38°C/W
Junction to Case ( $\theta_{JC}$ )	1°C/W
<b>Thermal Resistance, Four-Layer Board:</b>	
Junction to Ambient ( $\theta_{JA}$ )	28°C/W
Junction to Case ( $\theta_{JC}$ )	1°C/W

For the latest package outline information and land patterns (footprints), go to [www.maximintegrated.com/packages](http://www.maximintegrated.com/packages). Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to [www.maximintegrated.com/thermal-tutorial](http://www.maximintegrated.com/thermal-tutorial).

## Electrical Characteristics

( $V_M = +36V$ ,  $R_{ROFF} =$  from 15k $\Omega$  to 120k $\Omega$ ,  $R_{REF} =$  from 12k $\Omega$  to 72k $\Omega$ , Limits are 100% tested at  $T_A = +25^\circ C$ . Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Specifications marked "GBD" are guaranteed by design and not production tested.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
<b>POWER SUPPLY</b>						
Supply Voltage Range	$V_M$		4.5		65	V
Sleep Mode Current consumption	$I_{VM}$	$\overline{SLEEP} =$ logic low			20	$\mu A$
Quiescent Current Consumption	$I_{VM}$	$\overline{SLEEP} =$ logic high			5	mA
1.8V Regulator Output Voltage	$V_{VDD}$	$V_M = +4.5V$ , $I_{LOAD} = 20mA$		1.8		V
$V_{DD}$ Current Limit	$I_{VDD(LIM)}$	$V_{DD}$ shorted to GND	18			mA
Charge Pump Voltage	$V_{CP}$			$V_M + 2.7$		V

**Electrical Characteristics (continued)**

( $V_M = +36V$ ,  $R_{ROFF} =$  from 15k $\Omega$  to 120k $\Omega$ ,  $R_{REF} =$  from 12k $\Omega$  to 72k $\Omega$ , Limits are 100% tested at  $T_A = +25^\circ C$ . Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Specifications marked "GBD" are guaranteed by design and not production tested.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
<b>LOGIC LEVEL INPUTS-OUTPUTS</b>						
Input Voltage Level - High	$V_{IH}$		1.2			V
Input Voltage Level - Low	$V_{IL}$				0.65	V
Input Hysteresis	$V_{HYS}$			110		mV
Pulldown Current	$I_{PD}$	Logic supply ( $V_L$ ) = +3.3V	16	34	60	$\mu A$
Open-Drain Output Logic-Low Voltage	$V_{OL}$	$I_{LOAD} = 5mA$			0.4	V
Open-Drain Output Logic-High Leakage Current	$I_{OH}$	$V_{PIN} = +3.3V$	-1		1	$\mu A$
SLEEP Voltage Level High	$V_{IH(SLEEP)}$		0.9			V
SLEEP Voltage Level Low	$V_{IL(SLEEP)}$				0.6	V
SLEEP Pulldown Input Resistance	$R_{PD(SLEEP)}$		0.8	1.5		M $\Omega$
<b>OUTPUT SPECIFICATIONS</b>						
Output ON-Resistance Low Side	$R_{ON(LS)}$			150	270	m $\Omega$
Output ON-Resistance High Side	$R_{ON(HS)}$			150	300	m $\Omega$
Output Leakage	$I_{LEAK}$	Driver OFF	-12		12	$\mu A$
Dead Time	$t_{DEAD}$			100		ns
Output Slew Rate	SR			300		V/ $\mu s$
<b>PROTECTION CIRCUITS</b>						
Overcurrent Protection Threshold	OCP		3.8			A
Overcurrent Protection Blanking Time	$t_{OCP}$			2.2	3.5	$\mu s$
Autoretry OCP Time	$t_{RETRY}$			3		ms
UVLO Threshold on $V_M$	UVLO	$V_M$ rising	3.75	4	4.25	V
UVLO Threshold on $V_M$ Hysteris	UVLO <sub>HYS</sub>			0.12		V
Thermal Protection Threshold Temperature	$T_{SD}$			155		$^\circ C$
Thermal Protection Temperature Hysteresis	$T_{SD\_HYST}$			20		$^\circ C$
<b>CURRENT REGULATION</b>						
REF_ Pin Resistor Range	$R_{REF}$		12		72	K $\Omega$

**Electrical Characteristics (continued)**

( $V_M = +36V$ ,  $R_{ROFF} =$  from 15k $\Omega$  to 120k $\Omega$ ,  $R_{REF} =$  from 12k $\Omega$  to 72k $\Omega$ , Limits are 100% tested at  $T_A = +25^\circ C$ . Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Specifications marked "GBD" are guaranteed by design and not production tested.)

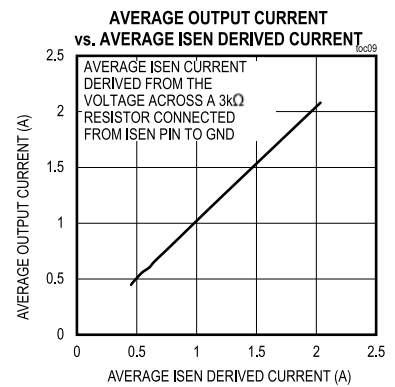
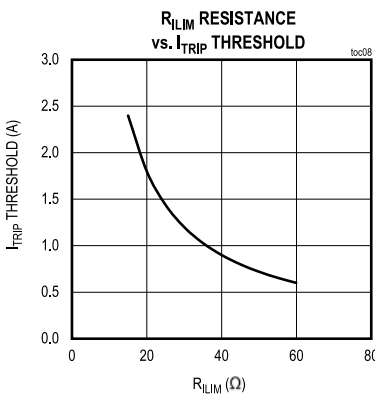
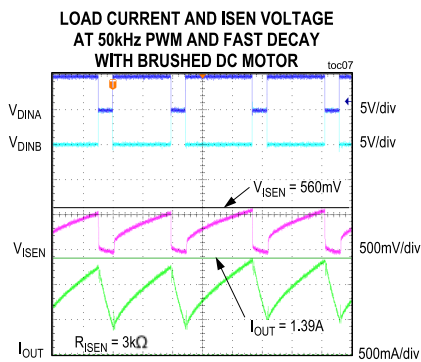
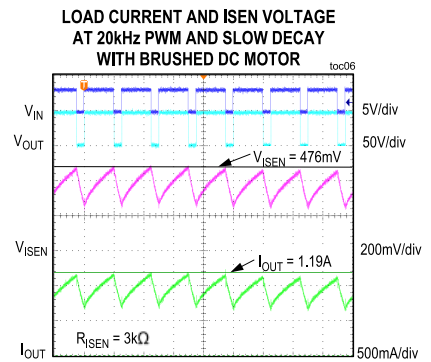
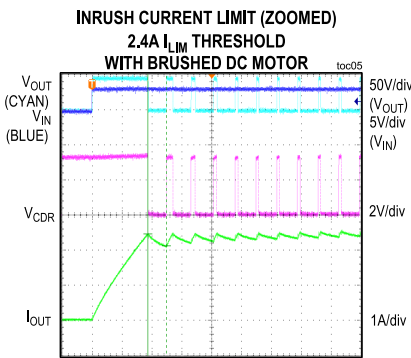
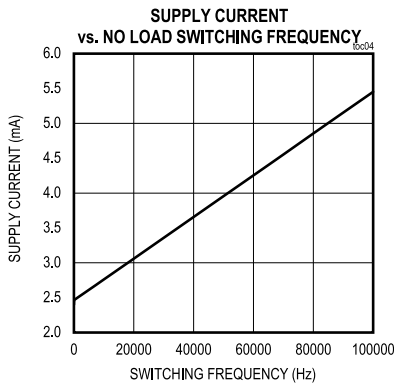
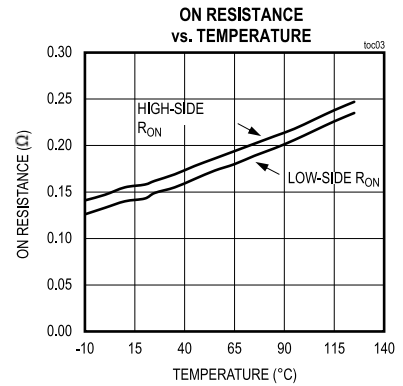
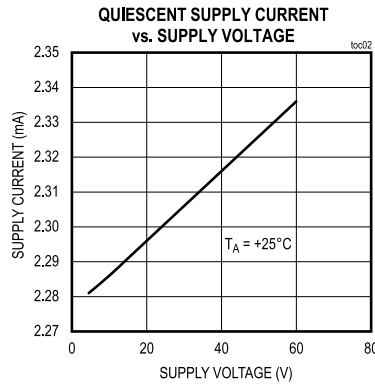
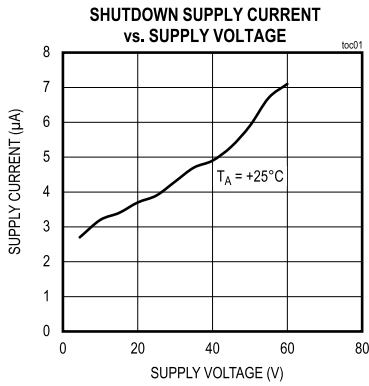
PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
REF Output Voltage	$V_{REF}$			900		mV
ITRIP Current Regulation Constant	KI			36		KV
Current Trip Regulation Accuracy ( <i>Note 1</i> )	DITRIP1	$I_{TRIP}$ from 1.75A to 3A	-5		5	%
	DITRIP2	$I_{TRIP}$ from 500mA to 1.75A	-10		+10	
Fixed OFF – Time Internal	$t_{OFF}$	$R_{OFF}$ shorted to $V_{DD}$	16	20	24	$\mu s$
Fixed OFF – Time Constant	KTOFF	$R_{ROFF}$ from 15K $\Omega$ to 120K $\Omega$		0.667		$\mu s/k\Omega$
PWM Blanking time	$t_{BLK}$			2.5		$\mu s$
<b>CURRENT SENSE MONITOR</b>						
ISEN_ Voltage Range	ISEN	Voltage Range at Pin ISEN	0		1.1	V
Current Monitor Scaling Factor	KISEN	Refer to the ISEN Output Current Equation in the <a href="#">Current Sense Output (CSO) - Current Monitor</a> Section		7500		A/A
Current Monitor Accuracy ( <i>Note 1</i> )	DKISEN1	$I_{OUT}$ from 1.1A to 3A	-5		+5	%
	DKISEN2	$I_{OUT}$ from 500mA to 1.1A	-10		+10	
Current Monitor Accuracy	DKISEN3	$I_{OUT}$ from 250mA to 500mA	-15		+15	%
Settling Time	$t_s$	$I_{FS} = I_{MAX}$		0.5		$\mu s$
<b>FUNCTIONAL TIMINGS</b>						
Sleep Time	$t_{SLEEP}$	$\overline{SLEEP} = 1$ to $OUT\_tristate$		40		$\mu s$
Wakeup Time From Sleep	$t_{WAKE}$	$\overline{SLEEP} = 0$ to normal operation			2.7	ms
Enable Time	$t_{EN}$	Time from EN pin rising edge to driver on			0.6	$\mu s$
Disable Time	$t_{DIS}$	Time from EN pin falling edge to driver off			1.4	$\mu s$

**Note 1:** Guaranteed by design, not production tested.



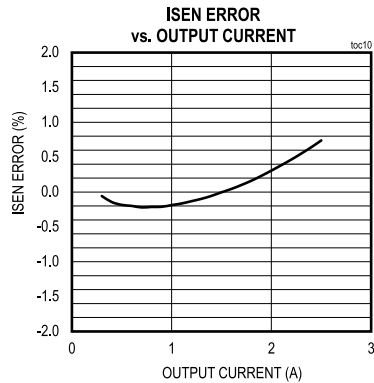
Typical Operating Characteristics

( $V_M = +4.5V$  TO  $+60V$ ;  $T_A = 25^\circ C$  unless otherwise noted.)



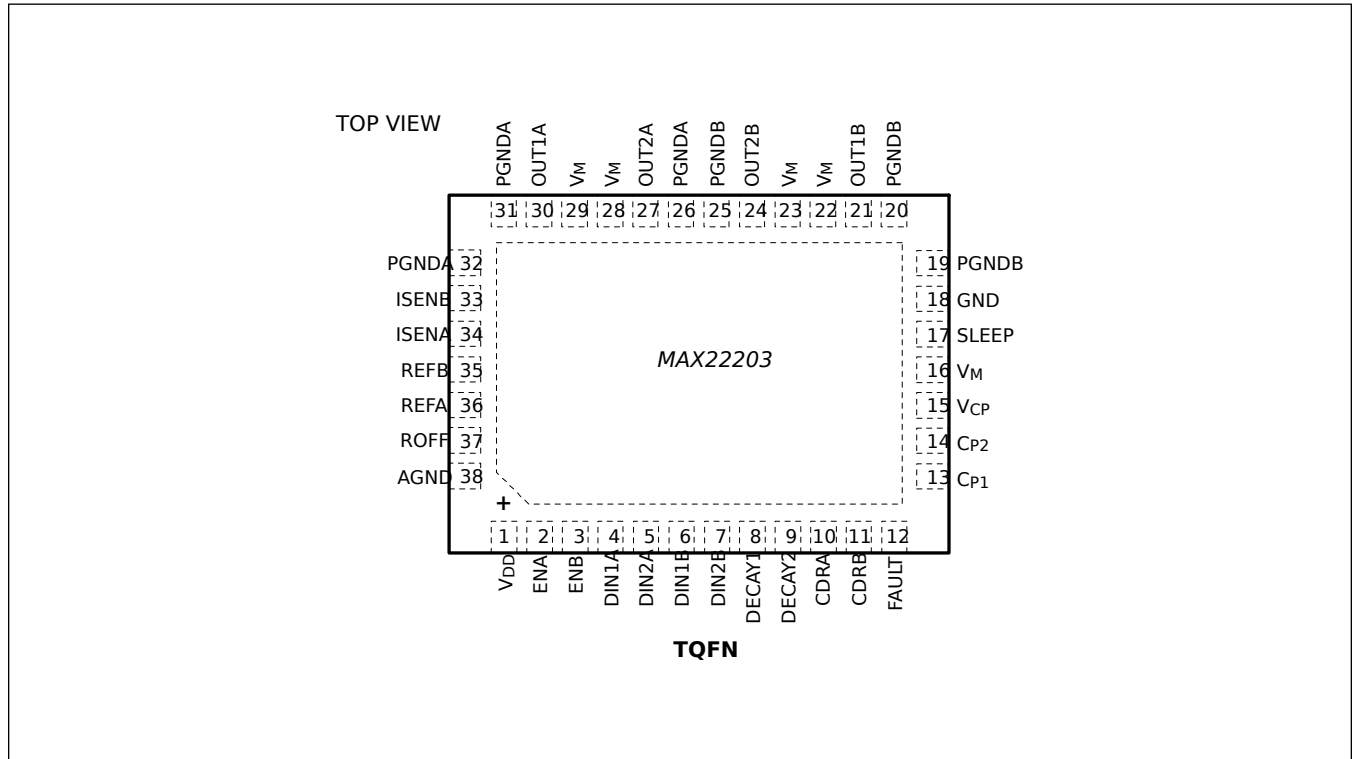
Typical Operating Characteristics (continued)

( $V_M = +4.5V$  TO  $+60V$ ;  $T_A = 25^\circ C$  unless otherwise noted.)



Pin Configuration

Pin Configuration



## Pin Description

PIN	NAME	FUNCTION	TYPE
16, 22, 23, 28, 29	V <sub>M</sub>	Supply Voltage Input. Connect at least 1 $\mu$ F SMD plus 10 $\mu$ F electrolytic bypass capacitors to GND. Higher values can be considered depending on application requirements.	Supply
15	V <sub>CP</sub>	Charge Pump Output. Connect a 5V, 1 $\mu$ F capacitor between V <sub>CP</sub> and V <sub>M</sub> as close as possible to the device	Output
13	C <sub>P1</sub>	Charge Pump Flying Capacitor Pin1. Connect a V <sub>M</sub> -rated 22nF capacitor between C <sub>P1</sub> C <sub>P2</sub> as close as possible to the device.	Output
14	C <sub>P2</sub>	Charge Pump Flying Capacitor Pin 2. Connect a V <sub>M</sub> -rated 22nF capacitor between C <sub>P1</sub> C <sub>P2</sub> as close as possible to the device.	Output
1	V <sub>DD</sub>	1.8V LDO Output. Connect a 5V, 2.2 $\mu$ F to GND close to the device	Analog Output
17	SLEEP	Active Low Sleep Pin	Logic Input
21, 24, 27, 30	OUT_	Driver Output Pins.	Output
12	FAULT	Open Drain Output Active Low Fault Indicator. Connect a 2K $\Omega$ resistor to the controller supply voltage.	Open Drain Output
33, 34	I <sub>SEN</sub> _	Current Sense Output Monitor. Connect a resistor to GND (ref. Current sense Output Detailed Description)	Output
2,3	EN_	Logic Input Pin. Enable Pin	Logic Input
4, 5, 6, 7	DIN_	CMOS PWM Input	Logic Input
8, 9	DECAY_	Logic Input. Set the Decay Mode	Logic Input
10, 11	CDR_	Open-Drain Output - Current Drive Regulator. Add a pullup resistor to the controller supply voltage. The pullup resistor value depends on the application requirements. Values between 1K $\Omega$ to 5K $\Omega$ meet the requirements for a majority of applications.	Open Drain Output
36	REFA	Programmable Current Analog Input. Connect a resistor from REFA to GND to set the current regulation threshold for Full Bridge A.	Analog Input
35	REFB	Programmable Current Analog Input. Connect a resistor from REFB to GND to set the current regulation threshold for Full Bridge B.	Analog Input
37	ROFF	t <sub>OFF</sub> Programmable Off Time Pin. Connect R <sub>OFF</sub> to V <sub>DD</sub> to use the internal fixed t <sub>OFF</sub> time. Connect a resistor from R <sub>OFF</sub> to GND to set the fixed OFF time to a desired value.	Analog Input
18, 38	GND	Analog Ground. Connect to Ground Plane.	GND
19, 20, 25, 26, 31, 32	PGND	Power GND. Connect to GND ground plane.	GND
EP	EP	Exposed PAD. Connect to GND.	GND

## Detailed Description

The MAX22203 is a Dual 65V, 3.8A<sub>MAX</sub> H-Bridge. It can be used to drive two Brushed DC motors or a Single Stepper Motor. The H-Bridge FETs have very low impedance, resulting in high driving efficiency and low heat generated. The typical total R<sub>ON</sub> (high-side + lowside) is 0.3Ω. Each H-Bridge can be individually PWM controlled by means of three logic inputs (DIN1, DIN2, and EN).

The MAX22203 features an accurate Current Drive Regulation (CDR), which can be used to limit the start up current of a Brushed DC motor or to control the phase current for stepper operation. The bridge output current is sensed by a non-dissipative Integrated Current Sensing (ICS) and it is then compared with the desired threshold current. As soon as the bridge current exceeds the threshold I<sub>TRIP</sub>, the device enforces the decay for a fixed OFF time (t<sub>OFF</sub>).

The non-dissipative ICS eliminates the bulky external power resistors, which are normally required for this function, resulting in a dramatic space and power saving compared with mainstream applications based on the external sense resistor.

A current proportional to the internally sensed motor current is output to an external pin (ISEN). By connecting an external resistor to this pin, a voltage proportional to the motor current is generated. The voltage built up on such external resistor can be input into the controller ADC whenever the motor control algorithm requires the current/torque information.

Also, two open-drain output pins (C<sub>DRA</sub>, C<sub>DRB</sub>) are asserted every time the internal current regulation takes control of the driver. This allows the external controller to monitor the activity of the internal current loop.

The maximum output current per H-Bridge is I<sub>MAX</sub> = 3.8A<sub>MAX</sub> and is limited by the Overcurrent Protection (OCP) circuit. This current can be driven for very short transients and is aimed to effectively drive small capacitive loads.

The maximum user configurable current regulation threshold is I<sub>TRIP\_MAX</sub> = 3A. Current thresholds can be set independently for the two full bridges by connecting the external resistors to pins R<sub>EFA</sub> and R<sub>EFB</sub>.

The maximum RMS current per H-Bridge is I<sub>RMS</sub> = 2A<sub>RMS</sub> on a standard JEDEC 4-layer board. Since this current is limited by thermal considerations, the actual maximum RMS current depends on the thermal characteristic of the application (PCB ground planes, heatsinks, forced air ventilation, etc).

The MAX22203 features Overcurrent Protection (OCP), Thermal Shutdown (TSD), and Undervoltage Lockout (UVLO). An open-drain active low FAULT pin is activated every time a fault condition is detected.

During Thermal shutdown and Undervoltage Lockout, the driver is tristated until normal operations are restored.

## Sleep Mode ( $\overline{\text{SLEEP}}$ Pin)

Drive this pin low to enter in the lowest power mode. All outputs are tristated and the internal circuits are biased off. The charge pump is also disabled. A pull-down resistor is connected between SLEEP and GND to ensure the part is disabled whenever this pin is not actively driven. This mode corresponds to the lowest power consumption possible. Waking up from Sleep Mode to Normal Mode takes up to 2.7ms max.

## PWM Control

When an H-Bridge is Enabled (EN<sub>+</sub> = Logic High) and the H-Bridge current is below the configured current limit, the average output voltage can be controlled by DIN1<sub>+</sub> and DIN1<sub>-</sub> logic input pins using PWM techniques. Setting Enable logic low causes the Output to enter a high impedance mode and the motor to coast. The Enable input pin frequency must not exceed 1KHz and cannot be used for PWM control.

[Table 1](#) shows the control Truth Table.

**Table 1. MAX22203 Truth Table**

EN <sub>+</sub>	DIN1 <sub>+</sub>	DIN2 <sub>+</sub>	OUT1	OUT2	DESCRIPTION
0	X	X	High-Z	High-Z	H bridge disabled. High impedance (HiZ)
1	0	0	L	L	Brake Low; Slow decay
1	1	0	H	L	Reverse (Current from OUT2 to OUT1)
1	0	1	L	H	Forward (Current from OUT1 to OUT2)

**Table 1. MAX22203 Truth Table (continued)**

1	1	1	H	H	Brake High; Slow Decay
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PWM techniques can be used to control the output duty cycle and hence to implement motor speed control. Typically, for brushed DC motor drivers, Slow Decay is preferred as it results in less ripple and higher efficiency. With this approach, during the OFF phase, both the low side FETs are activated effectively grounding the motor winding terminals. The current built up into the motor winding slowly decays. This decay is often referred to as Slow Decay. Alternatively, Fast Decay can also be implemented by reversing the bridge during the OFF phase.

### Current-Sense Output (CSO) - Current Monitor

Currents proportional to the internally sensed motor currents are output to pins ISENA and ISENB for H-bridge A and B respectively. The current is sensed when one of the two low side FETs sinks the output current and it is therefore meaningful for both during the energizing ( $t_{ON}$ ) phase and during the Slow Decay phase (Brake). In Fast Decay, the current is not monitored and ISEN outputs a zero current. The following equation shows the relationship between the current sourced at ISEN and the output current.

$$I_{ISEN}(A) = \frac{I_{OUT}(A)}{K_{ISEN}}$$

Equation - ISEN Output Current

in which  $K_{ISEN}$  represents the current scaling factor between the output current and its replica at pin ISEN.  $K_{ISEN}$  is typically 7500 A/A. For instance, if the instantaneous output current is 2A, the current sourced at ISEN is 266 $\mu$ A.

Figure [Figure 1](#) shows an idealized behavior of the ISEN current when Slow or Fast Decay are used. Blanking times, delays, and rise/fall edges have been ignored.

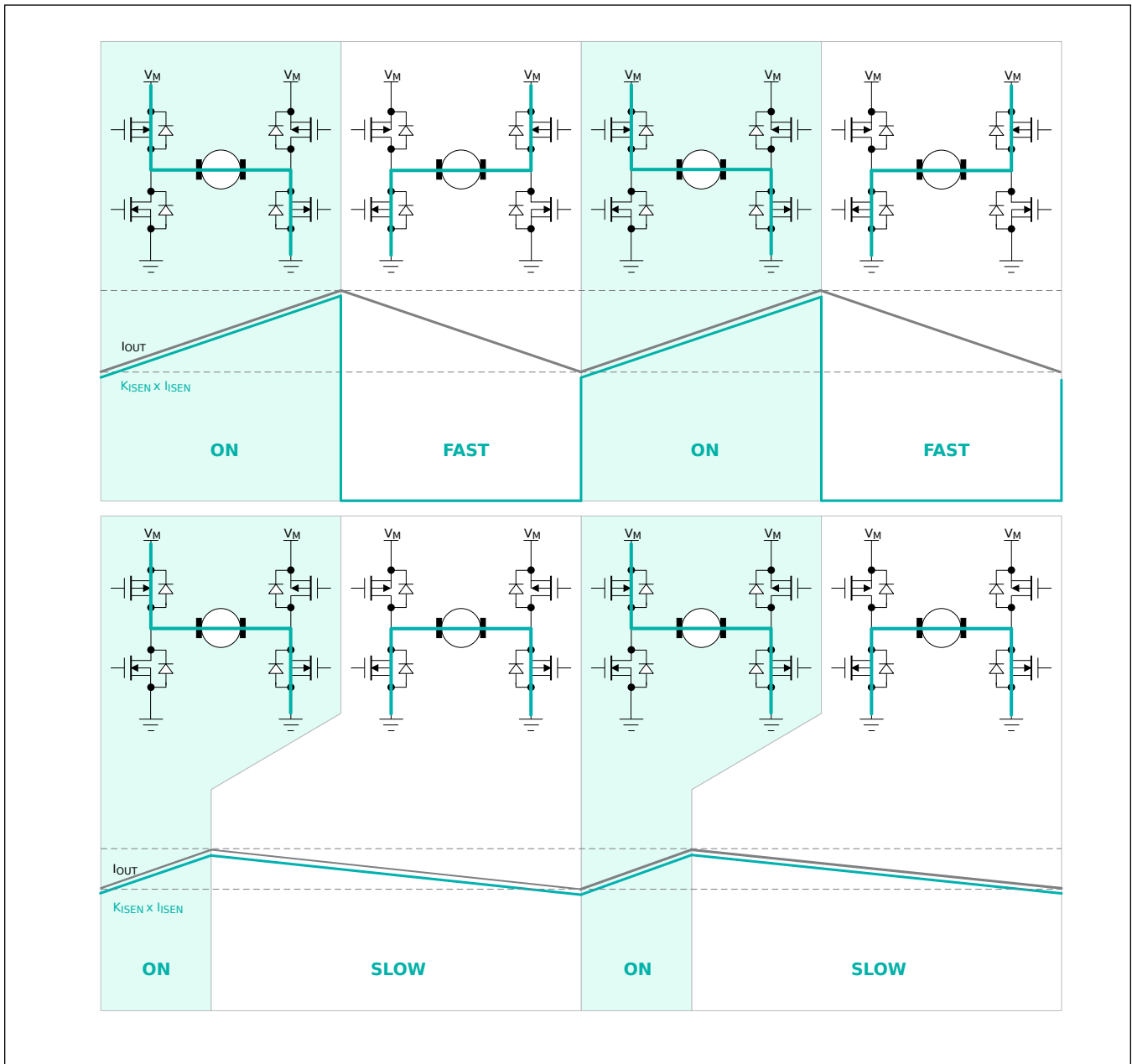


Figure 1. ISEN Current

By connecting an external signal resistor,  $R_{ISEN}$ , between ISEN and GND a voltage proportional to the motor current is generated. The voltage built up on  $R_{ISEN}$  can be input into the ADC of an external controller in applications in which the motor control algorithm requires the current/torque information. The following equation shows the design formula to calculate  $R_{ISEN}$  once the ADC full scale voltage ( $V_{FS}$ ) and the maximum operating current ( $I_{MAX}$ ) is known.

$$R_{ISEN}(\Omega) = K_{ISEN} \times \frac{V_{FS}(V)}{I_{MAX}(A)}$$

## Equation - RISEN Setting

For example, if the ADC operates up to 1V FS and the maximum operating output current is 2A, then  $R_{ISEN}$  is  $7500 \times 1V/2A = 3.75K\Omega$ .

The  $R_{ISEN}$  value also sets the output impedance of the Current-Sense Output circuit (ISEN output impedance). Normally, the input impedance of the ADC is much higher than  $R_{ISEN}$  enabling a direct connection to the ISEN pin without attenuation. In case a low input impedance ADC is used, a preamplifier (buffer) is required.

The Current-Sense Output circuit bandwidth and step response performances (see Specifications) ensure the current monitor tracks the driver current in motor drive applications.

**Current Drive Regulation**

The MAX22203 features embedded Current Drive Regulation (CDR).

The embedded current drive regulation provides an accurate control of the current flowing into the motor windings.

The bridge current is sensed by a non-dissipative Integrated Current Sensing circuit (ICS) and it is then compared with the threshold current ( $I_{TRIP}$ ). As soon as the bridge current exceeds the threshold, the device enforces the decay for a fixed OFF-time ( $t_{OFF}$ ). The device supports different decay modes as described in the following paragraphs.

Once  $t_{OFF}$  has elapsed, the driver is re-enabled for the next PWM cycle. During current regulation, the PWM duty cycle and frequency depend on the supply voltage, on the motor inductance, and on motor speed and load conditions.

The  $t_{OFF}$  duration can be configured with an external resistor connected to the ROFF pin.

**Integrated Current Sense (ICS)**

A non-dissipative Current Sensing is integrated. This feature eliminates the bulky external power resistors normally required for this function. This feature results in a dramatic space and power saving compared with mainstream applications based on the external sense resistor.

**Setting the Current Regulation Threshold – Pin REF**

Connect resistors from REFA and REFB to GND to set the current regulation thresholds for Full Bridge A and Full Bridge B respectively ( $I_{TRIPA}$ ,  $I_{TRIPB}$ ).

The equation below shows the typical  $I_{TRIP}$  current as a function of the  $R_{REF}$  shunt resistor connected to pin REF\_. The proportionality constant  $K_I$  is typically 36KV. The external resistor  $R_{REF}$  can range between 12K $\Omega$  and 72K $\Omega$ , which corresponds to  $I_{TRIP}$  setting ranging from about 3A down to 0.5A.

$$I_{TRIP} = \frac{K_I(KV)}{R_{REF}(K\Omega)}$$

**Setting the Fixed OFF\_TIME ( $t_{OFF}$ )**

The current regulation circuit is based on a constant  $t_{OFF}$  PWM control. When the bridge current exceeds the target  $I_{TRIP}$  current, an OFF phase begins and Decay modes are activated. The OFF phase has a fixed time duration ( $t_{OFF}$ ).  $t_{OFF}$  can be configured to a desired value by connecting an external resistor ( $R_{ROFF}$ ) to pin ROFF. When the ROFF pin is shorted to  $V_{DD}$ , the  $t_{OFF}$  time is internally set at a fixed value (20 $\mu$ s typical).

By connecting an external resistor to the pin  $R_{ROFF}$ , the user can configure  $t_{OFF}$  as shown in the equation below in which  $R_{ROFF}$  is an external resistor connected to the ROFF pin (in K $\Omega$ ) and  $K_{TOFF}$  is an internal constant equal to 0.667 $\mu$ s/K $\Omega$ .

$$t_{OFF}(\mu s) = R_{ROFF} \times K_{TOFF}$$

$t_{OFF}$  can be programmed from a range of 10 $\mu$ s to 80 $\mu$ s.

**CDR Open-Drain Output**

The CDR\_ pins are active-low open-drain outputs, which are asserted during the fixed  $t_{OFF}$  decay interval enforced by the integrated current drive regulation loop. An external controller monitoring the CDR\_ pins can determine whether the integrated current drive regulation loop has taken control of the driver overwriting the status of the PWM logic inputs (DIN1, DIN2).

The CDR\_ signals can be used by an external controller for a variety of reasons and provides information about the actual load during current regulation. For example, in the use case where the PWM are permanently held in Forward or Reverse mode, control of the motor current is entrusted to the internal Current Drive Regulation loop and the CDR\_ pin status directly reflects the driver output status. In this example, the duty cycle of the CDR\_ pin can be used to detect stall conditions.

A pullup resistor must be connected from the CDR\_ pins to the controller voltage supply. The pullup resistor choice depends on the PCB line capacitance, PWM frequency, and power consumption. Values between 1K $\Omega$  to 5K $\Omega$  satisfy the requirement for most applications.

The time diagram in [Figure 2](#) shows the behavior of this function when the motor spins in forward direction respectively with DIN2 held firmly High (Case A) or when DIN2 is toggling (Case B and C).

The CDR output is asserted only when the slow decay mode is forced by the internal CDR.

Notice that any PWM transitions resets the fixed OFF Time of the CDR circuit. In Case B, the actual Slow Decay Interval is longer than  $t_{OFF}$  whereas in Case C, the actual Slow Decay OFF interval is shorter.



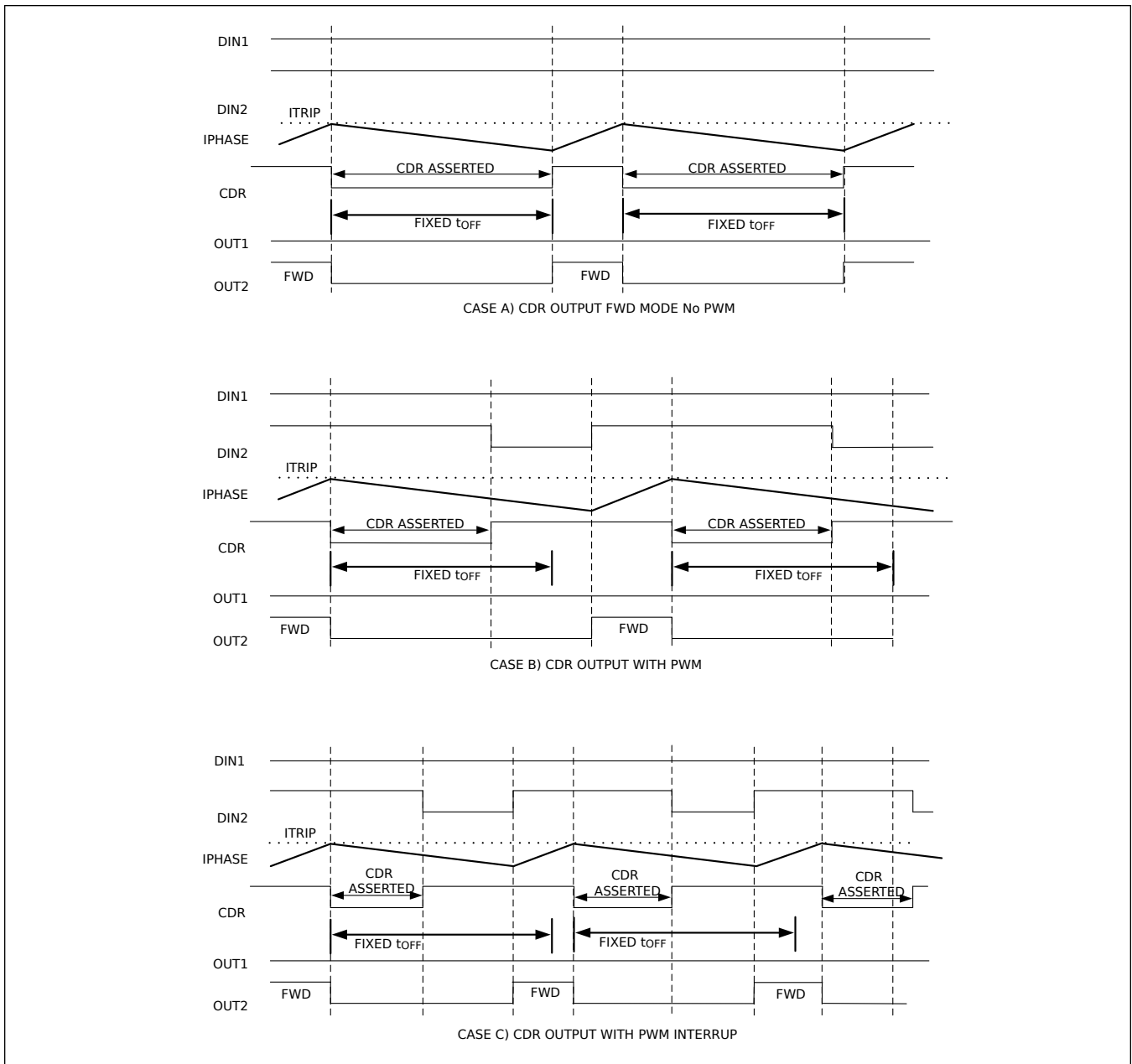


Figure 2. CDR Monitor Timing Diagram

**Operating Modes**

During PWM chopping, the driver output alternates Energizing (ON) and Decay phases. The MAX22203 supports different Decay modes. Slow Decay, Fast Decay, and different combination between Slow and Fast are supported.

Figure 3 shows the current path in the three different modes of operation.

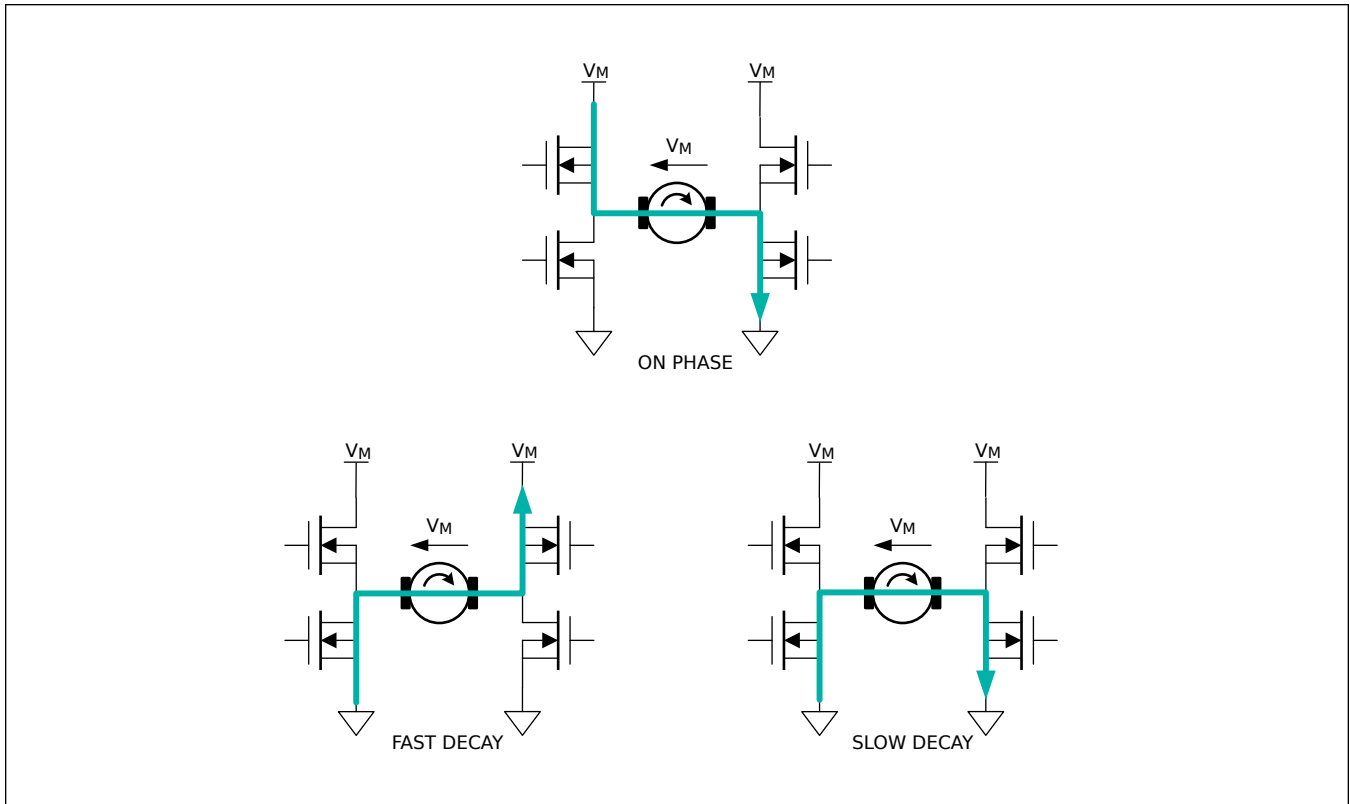


Figure 3. Current Flow During ON and Decay Modes

### Setting the Decay Mode

Two logic input pins allow the user to set the Decay Mode during  $t_{OFF}$ . The MAX22203 supports Slow, Fast, and Mixed Decay modes.

Table 2 shows the Truth Table for the Decay selection.

**Table 2. Decay Mode Truth Table**

DECAY2	DECAY1	DECAY MODE
0	0	SLOW
0	1	MIXED 30% FAST* / 70% SLOW
1	0	MIXED 60% FAST* / 40% SLOW
1	1	FAST*

\* To prevent reversal of current during fast decay, outputs go to the high-impedance state as the current approaches 0A.

### Protections

#### Overcurrent Protection (OCP)

An Overcurrent Protection (OCP) protects the device against short circuits to the rails (supply voltage and ground) and across the load terminals. The OCP threshold is set at 3.8A minimum. If the output current is larger than the OCP threshold for longer than the OCP blanking time, then an OCP event is detected.

When an OCP event is detected, the H-Bridge is immediately disabled, and a fault indication is output on the pin  $\overline{FAULT}$ . The H-Bridge is kept in a high impedance mode for 3ms (see  $t_{RETRY}$  specification). After that, the H-Bridge is re-enabled according to the current state. If the short circuit is still present, this cycle repeats. Otherwise, normal operation resumes.

It is recommended to avoid prolonged operation under the short-circuit failure mode since a prolonged OCP auto-retry could affect the device reliability.

**Thermal Shutdown Protection (TSD)**

If the die temperature exceeds 155°C (typical value), a fault indication is output on pin  $\overline{\text{FAULT}}$  and the driver is tri-stated until the junction temperature drops below 135°C. After that, the driver is re-enabled.

**Undervoltage Lockout Protection (UVLO)**

The device features Undervoltage Lockout Protection (UVLO). UVLO on  $V_M$  is set at 4.25V maximum. When an UVLO event occurs, a fault indication is output on pin  $\overline{\text{FAULT}}$  and the driver outputs are tristated. Normal operation is then resumed (and the  $\overline{\text{FAULT}}$  pin deasserted) as soon as the supply voltages are back in the nominal operating range.

Applications Information

Recommended Layout

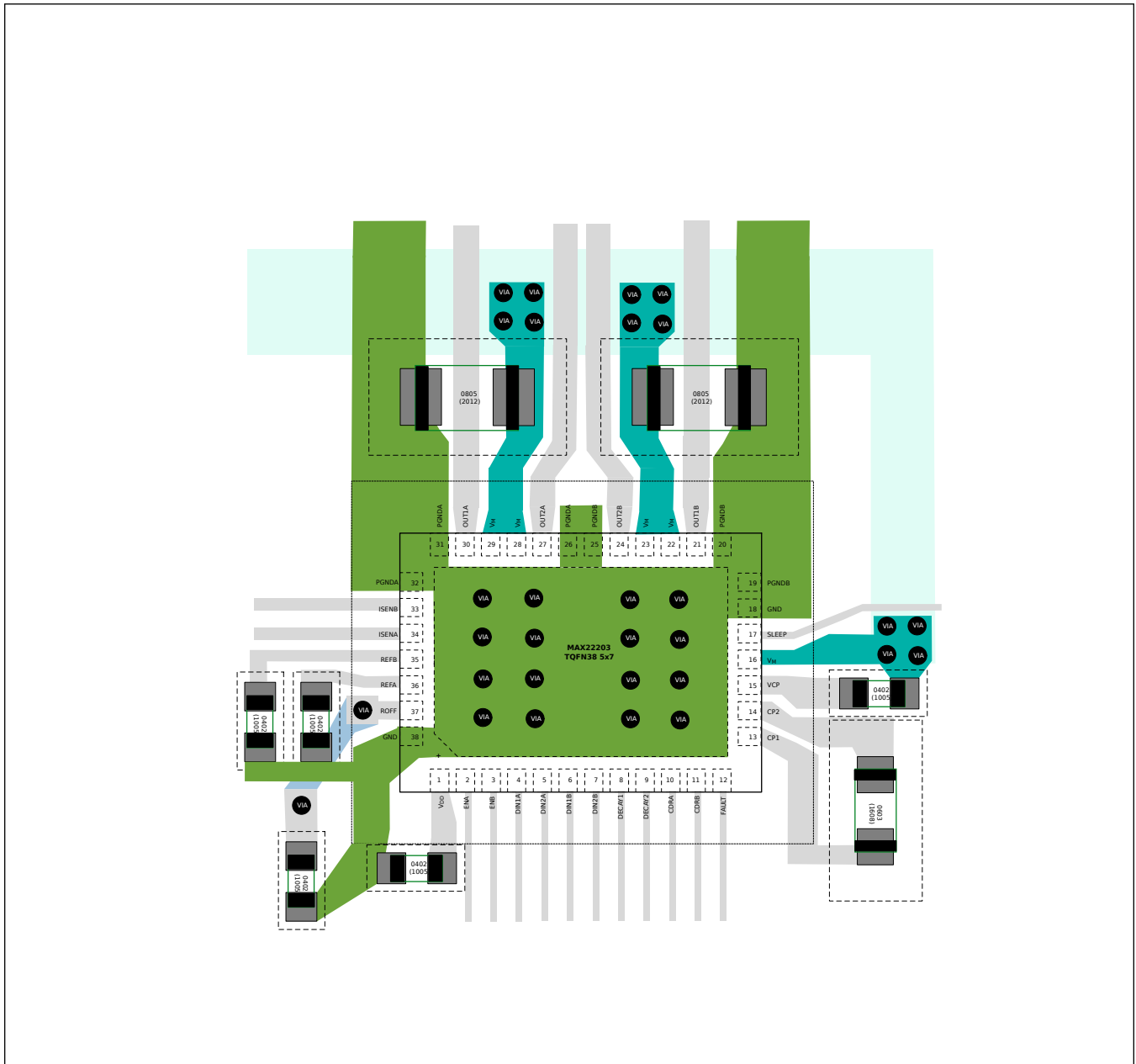
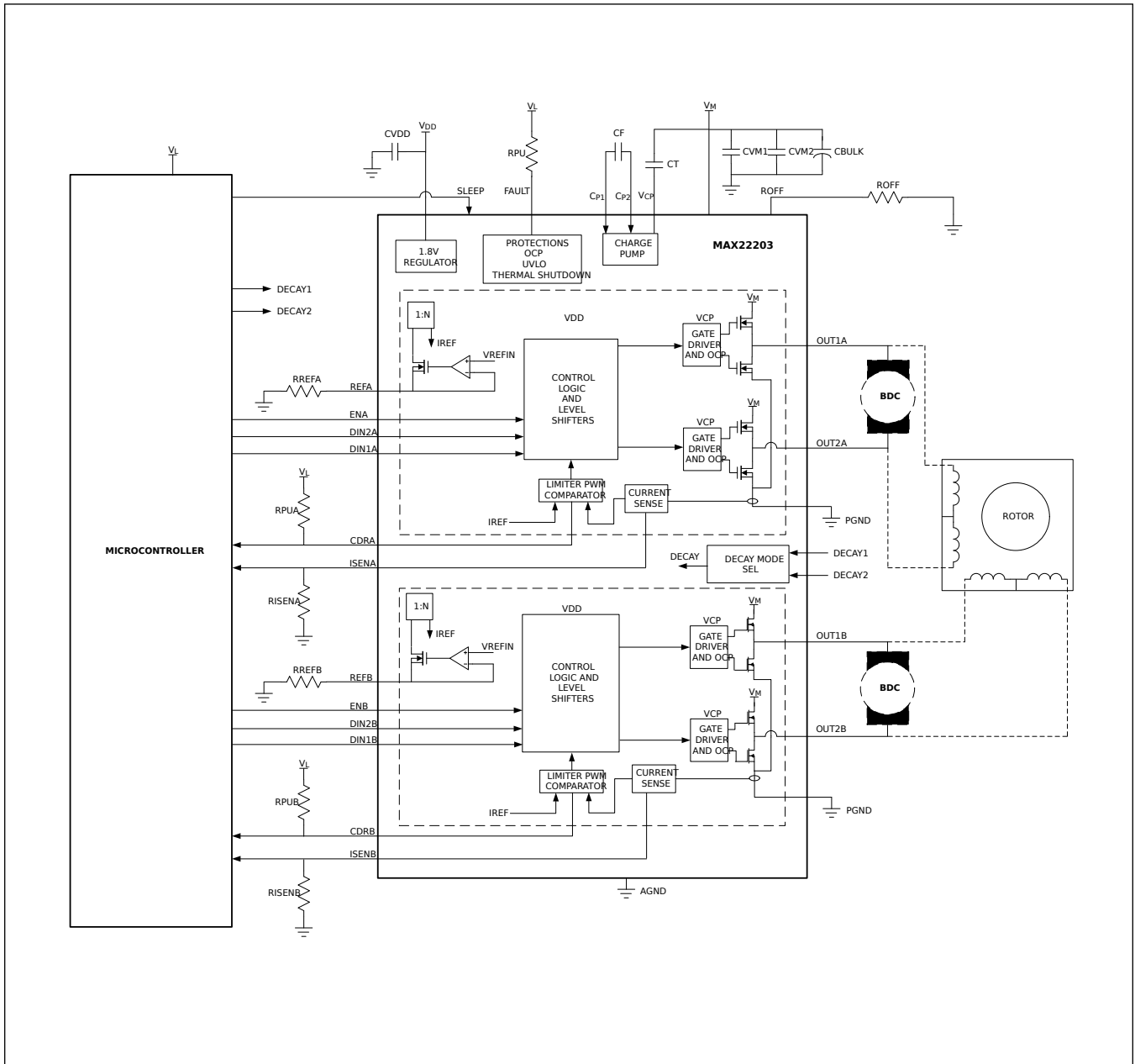


Figure 4. Recommended Layout

Typical Application Circuits

Application Diagram



MAX22203

65V, 3.8A Dual Brushed or Single Stepper Motor  
Driver with Integrated Current Sense

### Ordering Information

PART NUMBER	TEMPERATURE RANGE	PIN-PACKAGE
MAX22203ATU+	-40°C to +125°C	38 TQFN
MAX22203AHU+*	-40°C to +125°C	38 TSSOP

+ Denotes a lead(Pb)-free/RoHS-compliant package.

T Denotes tape-and-reel.

\* Denotes future product. Contact factory for availability.

MAX22203

65V, 3.8A Dual Brushed or Single Stepper Motor  
Driver with Integrated Current Sense

## Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	5/21	Initial release	—

For pricing, delivery, and ordering information, please visit Maxim Integrated's online storefront at <https://www.maximintegrated.com/en/storefront/storefront.html>.

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