MAX30001

Ultra-Low-Power, Single-Channel Integrated Biopotential (ECG, R-to-R, and Pace Detection) and Bioimpedance (BioZ) AFE

General Description

The MAX30001 is a complete, biopotential and bioimpedance (BioZ), analog front-end (AFE) solution for wearable applications. It offers high performance for clinical and fitness applications, with ultra-low power for long battery life. The MAX30001 is a single biopotential channel providing electrocardiogram (ECG) waveforms, heart rate and pacemaker edge detection, and a single bioimpedance channel capable of measuring respiration.

The biopotential and bioimpedance channels have ESD protection, EMI filtering, internal lead biasing, DC leadsoff detection, ultra-low-power, leads-on detection during standby mode, and extensive calibration voltages for builtin self-test. Soft power-up sequencing ensures no large transients are injected into the electrodes. Both channels also have high input impedance, low noise, high CMRR, programmable gain, various low-pass and high-pass filter options, and a high resolution analog-to-digital converter. The biopotential channel is DC coupled, can handle large electrode voltage offsets, and has a fast recovery mode to quickly recover from overdrive conditions, such as defibrillation and electro-surgery. The bioimpedance channel includes integrated programmable current drive, works with common electrodes, and has the flexibility for 2 or 4 electrode measurements. The bioimpedance channel also has AC lead off detection.

The MAX30001 is available in a 30-bump wafer-level package (WLP), operating over the 0°C to +70°C commercial temperature range.

Applications

- Single-Lead Event Monitors for Arrhythmia Detection
- Single-Lead Wireless Patches for In-Patient/Out-Patient Monitoring
- Chest Band Heart Rate Monitors for Fitness Applications
- Bio Authentication and ECG-On-Demand Applications
- Respiration and Hydration Monitors
- Impedance Based Heart Rate Detection

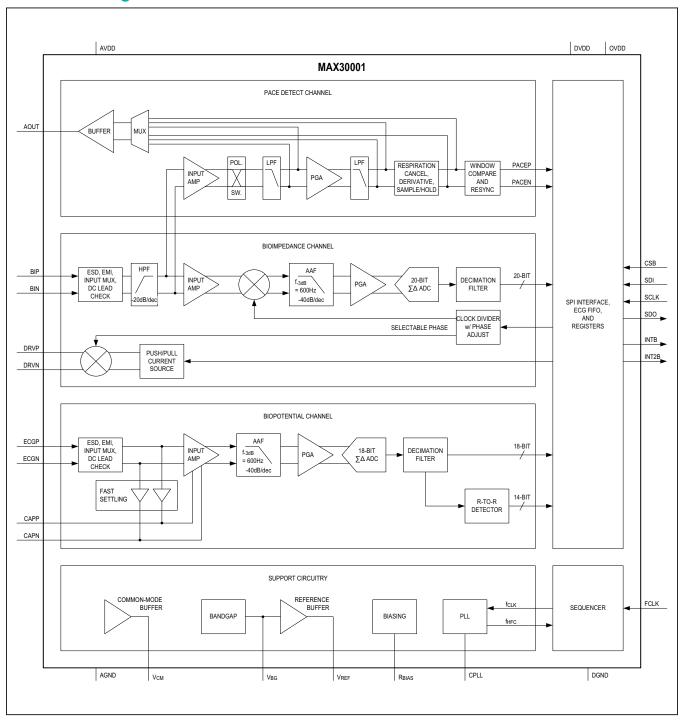
Ordering Information appears at end of data sheet.

Benefits and Features

- Can be Used in IEC 60601-2-47:2012 Compliant Systems
- Clinical-Grade ECG and BioZ AFE with High Resolution Data Converter
 - 15.9 Bits ENOB with 3.1µV_{P-P} (typ) Noise for ECG
 - 17 Bits ENOB with 1.1μV_{P-P} Noise for BioZ
- Better Dry Starts Due to Much Improved Real World CMRR and High Input Impedance
 - Fully Differential Input Structure with CMRR > 100dB
- Offers Better Common-Mode to Differential Mode Conversion Due to High Input Impedance
- High Input Impedance > 1GΩ for Extremely Low Common-to-Differential Mode
- Minimum Signal Attenuation at the Input During Dry Start Due to High Electrode Impedance
- High DC Offset Range of ±650mV (1.8V, typ) Allows to Be Used with Wide Variety of Electrodes
- High AC Dynamic Range of 65mV_{P-P} for ECG and 90mV_{P-P} for BioZ Will Help Prevent Saturation in the Presence of Motion/Direct Electrode Hits
- Longer Battery Life Compared to Competing Solutions
 - 85µW at 1.1V Supply Voltage for ECG
 - 158µW at 1.1V Supply Voltage for BioZ
- Leads-On Interrupt Feature Allows to Keep the µC in Deep Sleep Mode Until Valid Lead Condition is Detected
 - Lead-On Detect Current: 0.7µA (typ)
- Built-In Heart Rate Detection with Interrupt Feature Eliminates the Need to Run HR Algorithm on the µController
 - Robust R-R Detection in High Motion Environment at Extremely Low Power
- Configurable Interrupts Allows the μC Wake-Up Only on Every Heart Beat Reducing the Overall System Power
- High Accuracy Allows for More Physiological Data Extractions
- 32-Word ECG and 8-Word BioZ FIFOs Allows the MCU to Stay Powered Down for 256ms with Full Data Acquisition
- High-Speed SPI Interface
- Shutdown Current of 0.6µA (typ)



Functional Diagram



MAX30001

Ultra-Low-Power, Single-Channel Integrated Biopotential (ECG, R-to-R, and Pace Detection) and Bioimpedance (BioZ) AFE

Absolute Maximum Ratings

AVDD to AGND0.3V to +2.0V	Maximum Current into Any Pin±50mA
DVDD to DGND0.3V to +2.0V	Continuous Power Dissipation (T _A = +70°C)
AVDD to DVDD0.3V to +0.3V	30-Bump WLP
OVDD to DGND0.3V to +3.6V	(derate 24.3mW/°C above +70°C)1945.5mW
AGND to DGND0.3V to +0.3V	Operating Temperature Range0°C to +70°C
CSB, SCLK, SDI, FCLK to DGND0.3V to +3.6V	Junction Temperature+150°C
SDO, INTB, INT2B	Storage Temperature Range65°C to +150°C
to DGND0.3V to the lower of (3.6V and OVDD + 0.3V)	Lead Temperature (Soldering, 10sec)+300°C
All Other Pins	Soldering Temperature (reflow)+260°C
to AGND0.3V to the lower of (2.0V and AVDD + 0.3V)	

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Package Information

PACKAGE TYPE: 30 WLP					
Package Code	W302L2+1				
Outline Number	21-100074				
Land Pattern Number	Refer to Application Note 1891				
THERMAL RESISTANCE, FOUR-LAYER BOARD					
Junction to Ambient (θ _{JA})	44°C/W				

Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to www.maximintegrated.com/thermal-tutorial.

For the latest package outline information and land patterns (footprints), go to www.maximintegrated.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

Electrical Characteristics

 $(V_{DVDD} = V_{AVDD} = +1.1V \text{ to } +2.0V, V_{OVDD} = +1.65V \text{ to } +3.6V, f_{FCLK} = 32.768 \text{kHz}, LN_BIOZ = 1, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted.}$ Typical values are at $V_{DVDD} = V_{AVDD} = +1.8V, V_{OVDD} = +2.5V, T_A = +25^{\circ}C.)$ (Note 1)

PARAMETER	SYMBOL	CONDITIONS		TYP	MAX	UNITS
ECG CHANNEL						
AC Differential Input Pange		V _{AVDD} = +1.1V, THD < 0.3%	-15		+15	m\/
AC Differential Input Range		V _{AVDD} = +1.8V, THD < 0.3%		±32.5		mV _{P-P}
DC Differential Input Benge		V _{AVDD} = +1.1V, shift from nominal gain < 2%	-300		+300	m\/
DC Differential Input Range		V _{AVDD} = +1.8V		±650		mV
Common Mode Innut Dance		V_{AVDD} = +1.1V, from V_{MID} , shift from nominal gain < 2%	-150		+150	\/
Common Mode Input Range		V_{AVDD} = +1.8V, from V_{MID} , shift from nominal gain < 2%		±550		- mV
Common Mode Rejection Ratio	CMRR	0Ω source impedance, f = 64Hz, T_A = +25°C (Note 2)	100	115		dB
		With impedance mismatch (Note 3)		77		

Electrical Characteristics (continued)

 $(V_{DVDD} = V_{AVDD} = +1.1 \text{V to } +2.0 \text{V}, V_{OVDD} = +1.65 \text{V to } +3.6 \text{V}, f_{FCLK} = 32.768 \text{kHz}, LN_BIOZ = 1, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted.}$ Typical values are at $V_{DVDD} = V_{AVDD} = +1.8 \text{V}, V_{OVDD} = +2.5 \text{V}, T_A = +25 ^{\circ}\text{C}.)$ (Note 2)

PARAMETER	SYMBOL	CONDI	TIONS	MIN	TYP	MAX	UNITS
		DW = 0.05 450U= 0	- 20v		0.77		μV _{RMS}
ECG Channel Input Referred		BVV = 0.05 - 150HZ, GC	BW = 0.05 – 150Hz, G _{CH} = 20x		4.6		μV _{P-P}
Noise		DW - 0.05 40H- C	= 20v (Note 2)		0.46	1.0	μV _{RMS}
		BW = $0.05 - 40$ Hz, G _C H	1 = 20x (Note 2)		3.1	6.6	µV _{P-P}
Input Leakage Current		T _A = +25°C		-1	±0.1	+1	nA
Input Impadance (INIA)		Common-mode, DC			45		GΩ
Input Impedance (INA)		Differential, DC			1500		МΩ
ECG Channel Total Harmonic	THD	V _{AVDD} = +1.80V, V _{IN} = 6 G _{CH} = 20x, electrode off			0.025		- %
Distortion	IND	V_{AVDD} = +1.1V, V_{IN} = 30 G_{CH} = 20x, electrode of				0.3	70
ECG Channel Gain Setting	G _{CH}	Programmable, see EC	G_GAIN[1:0]		20 to 160		V/V
ECG Channel Gain Error		V _{AVDD} = +1.8V, G _{CH} = 2 ECGP = ECGN = VMID	20x,	-2.5		+2.5	%
(Excluding Reference)		V _{AVDD} = +1.1V, G _{CH} = 2 ECGP = ECGN = VMID	20x,	-4.5		+4.5	%
ECG Channel Offset Error		(Note 4)			±0.1		% of FSR
ADC Resolution					18		Bits
ADC Sample Rate		Programmable, see EC	G_RATE[1:0]		125 to 512		SPS
CAPP to CAPN Impedance	R _{HPF}	FHP = 1/(2π x R _{HPF} x C capacitance between C		320	450	600	kΩ
		Fast recovery enabled (1.8V)		160		
Analog High-Pass Filter Slew Current		Fast recovery enabled (1.1V)		55		μA
Current		Fast recovery disabled			0.09		1
Fast Settling Recovery Time		C _{HPF} = 10μF, Note: varies by sample rate, see Table 3.			500		ms
			DLPF[0:1] = 01		40		
Digital Low-Pass Filter		Linear phase FIR filter.	DLPF[0:1] = 10		100		Hz
		ECG_RATE = 00, 01			150		1
Digital High-Pass Filter		Phase-corrected 1st-ord	er IIR filter. DHPF = 1		0.5		Hz
ECC Dower Supply Dainetis	Denn	Lead bias disabled, DC			107		40
ECG Power Supply Rejection	PSRR	Lead bias disabled, f = 6	64Hz		110		dB

Electrical Characteristics (continued)

 $(V_{DVDD} = V_{AVDD} = +1.1 \text{V to } +2.0 \text{V}, V_{OVDD} = +1.65 \text{V to } +3.6 \text{V}, f_{FCLK} = 32.768 \text{kHz}, LN_BIOZ = 1, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted.}$ Typical values are at $V_{DVDD} = V_{AVDD} = +1.8 \text{V}, V_{OVDD} = +2.5 \text{V}, T_A = +25 ^{\circ}\text{C}.)$ (Note 2)

PARAMETER	SYMBOL	CONDI	TIONS	MIN TYP	MAX	UNITS
ECG INPUT MUX						1
			IMAG[2:0] = 001	5		
			IMAG[2:0] = 010	10		1
DC Lead Off Check		Pullup/ pulldown	IMAG[2:0] = 011	20		nA
		pulldown	IMAG[2:0] = 100	50		1
			IMAG[2:0] = 101	100		1
		VTH[1:0] = 11 (Note 5)		V _{MID} – 0.50		
DC Lead Off Comparator Low		VTH[1:0] = 10 (Note 6)		V _{MID} – 0.45		
Threshold		VTH[1:0] = 01 (Note 7)		V _{MID} – 0.40		V
		VTH[1:0] = 00		V _{MID} – 0.30		
		VTH[1:0] = 11 (Note 5)		V _{MID} + 0.50		
DC Lead Off Comparator High		VTH[1:0] = 10 (Note 6)		V _{MID} + 0.45		V
Threshold		VTH[1:0] = 01 (Note 7)		V _{MID} + 0.40		
		VTH[1:0] = 00		V _{MID} + 0.30		
			RBIASV[1:0] = 00	50		
Lead Bias Impedance		Lead bias enabled	RBIASV[1:0] = 01	100		ΜΩ
			RBIASV[1:0] = 10	200		1
Lead Bias Voltage	V _{MID}	Lead bias enabled		V _{AVDD} / 2.15		V
Calibration Valtage Magnitude		Cingle anded	V _{MAG} = 0	0.25		\/
Calibration Voltage Magnitude		Single-ended	V _{MAG} = 1	0.50		mV
Calibration Voltage Magnitude Error		Single-ended (Note 8)		-3	+3	%
Calibration Voltage Frequency		Programmable, see FC	AL[2:0]	0.0156 to 256		Hz
Calibration Voltage Pulse Time		Programmable, see	FIFTY = 0	0.03052 to 62.474		ms
		THIGH[10:0]	FIFTY = 1	50		%

Electrical Characteristics (continued)

 $(V_{DVDD} = V_{AVDD} = +1.1 \text{V to } +2.0 \text{V}, V_{OVDD} = +1.65 \text{V to } +3.6 \text{V}, f_{FCLK} = 32.768 \text{kHz}, LN_BIOZ = 1, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted.}$ Typical values are at $V_{DVDD} = V_{AVDD} = +1.8 \text{V}, V_{OVDD} = +2.5 \text{V}, T_A = +25 ^{\circ}\text{C}.)$ (Note 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
BIOIMPEDANCE (BioZ) CHANN	IEL			,		'
Signal Generator Resolution		Square wave generator		1		Bits
DRVP/N Injected Full-Scale Current		Programmable, see BIOZ_CGMAG[2:0]		8 to 96		μA _{PK}
DRVP/N Injected Current		Internal bias resistor, see EXT_RBIAS	-30		+30	- %
Accuracy		External bias resistor (0.1%, 10ppm, 324kΩ)	-10		+10	70
DRVP/N Injected Current Power Supply Rejection				<±1		%/V
DRVP/N Injected Current Temperatue Coefficient		External bias resistor, 32μA _{P-P} , 0 to 70°C (0.1%, 10ppm, 324kΩ)		50		ppm/°C
DRVP/N Compliance Voltage		V _{DRVP} - V _{DRVN}		±(V _{AVDD} - 0.5)		V _{P-P}
Current Injection Frequency		Programmable, see BIOZ_FCGEN[3:0]		0.125 to 131.072		kHz
AC Differential Input Penge		Shift from nominal gain < 1% (V _{AVDD} = 1.1V)		25		mV
AC Differential Input Range		Shift from nominal gain < 1% (V _{AVDD} = 1.8V)		90		mV
BioZ Channel Gain		Programmable, see BIOZ_GAIN[1:0]		10 to 80		V/V
ADC Sample Rate		Programmable, see BIOZ_RATE		24.98 to 64		sps
ADC Resolution				20		Bits
Input Referred Noise		BW = 0.05 to 4Hz, Gain = 20x		0.16		μV _{RMS}
(BIP, BIN)		BW = 0.05 to 4Hz, Gain = 20x		1.1		μV _{P-P}
Impedance Resolution		DC to 4Hz, $32\mu A_{P-P}$, $40kHz$, Gain = $20x$, R_{BODY} = 680Ω		40		mΩ _{P-P}
Input Analog High Pass Filter		Programmable, see BIOZ_AHPF[2:0]		125 to 7200		Hz
Demodulation Phase Range		Programmable, see BIOZ_PHOFF[3:0]		0 - 168.75		۰
Demodulation Phase Resolution				11.25		0
		BIOZ_DLPF[1:0] = 01		4		
Output Digital Low Pass Filter		BIOZ_DLPF[1:0] = 10		8		Hz
		BIOZ_DLPF[1:0] = 11		16		<u></u>
Output Digital High Pass Filter		BIOZ_DHPF[1:0] = 01		0.05		Hz
Output Digital High Fass Filler		BIOZ_DHPF[1:0] = 1x		0.5		Hz

Electrical Characteristics (continued)

 $(V_{DVDD} = V_{AVDD} = +1.1 \text{V to } +2.0 \text{V}, V_{OVDD} = +1.65 \text{V to } +3.6 \text{V}, f_{FCLK} = 32.768 \text{kHz}, LN_BIOZ = 1, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted.}$ Typical values are at $V_{DVDD} = V_{AVDD} = +1.8 \text{V}, V_{OVDD} = +2.5 \text{V}, T_A = +25 ^{\circ}\text{C}.)$ (Note 2)

PARAMETER	SYMBOL	CONDI	TIONS	MIN TYP	MAX	UNITS
BIOIMPEDANCE (BioZ) INPUT	MUX					
		IMAG[2:0] = 001		5		
		IMAG[2:0] = 010		10		1
DC Lead Off Check		IMAG[2:0] = 011		20		nA
		IMAG[2:0] = 100		50		1
		IMAG[2:0] = 101		100]
		DCLOFF_VTH[1:0] = 11	(Note 5)	V _{MID} - 0.5	0	
DC Lead Off Comparator Low		DCLOFF_VTH[1:0] = 10) (Note 6)	V _{MID} - 0.4	5	V
Threshold		DCLOFF_VTH[1:0] = 0 ⁻¹	1 (Note 7)	V _{MID} - 0.4	0] v
		DCLOFF_VTH[1:0] = 00)	V _{MID} - 0.3	0	
		DCLOFF_VTH[1:0] = 11	(Note 5)	V _{MID} + 0.5	0	
DC Lead Off Comparator High		DCLOFF_VTH[1:0] = 10) (Note 6)	V _{MID} + 0.4	5	V
Threshold		DCLOFF_VTH[1:0] = 0 ⁻¹	V _{MID} + 0.4	0] v	
		DCLOFF_VTH[1:0] = 00		V _{MID} + 0.3		
		Lead bias enabled, RBIASV[1:0] = 00 Lead bias enabled, RBIASV[1:0] = 01 Lead bias enabled, RBIASV[1:0] = 10		50		
Lead Bias Impedance				100	ΜΩ	
				200		
Lead Bias Voltage	V_{MID}	Lead bias enabled.		V _{AVDD} / 2.15		V
Calibration Valtage Magnitude		Single-ended. V _{MAG} = 0	0	0.25		mV
Calibration Voltage Magnitude		Single-ended. V _{MAG} =	1	0.50] 1111
Calibration Voltage Error		Single-ended. (Note 8)		-3	+3	%
Calibration Voltage Frequency		Programmable, see FC	AL[2:0]	0.0156 to 256	1	Hz
Calibration Voltage Pulse Time		Programmable,	FIFTY = 0	0.03052 to 62.474		ms
Ç		see THIGH[10:0]	FIFTY = 1	50		%
Resistive Load Nominal Value	R _{VAL}	Programmable, see BMUX_RNOM[2:0]		0.625 to 5	.0	kΩ
Resistive Load Modulation Value	R _{MOD}	Programmable, see BM	Programmable, see BMUX_RMOD[2:0]		0	mΩ
Resistive Load Modulation Frequency	F _{MOD}	Programmable, see BM	Programmable, see BMUX_FBIST[1:0]		.0	Hz

Electrical Characteristics (continued)

 $(V_{DVDD} = V_{AVDD} = +1.1 \text{V to } +2.0 \text{V}, V_{OVDD} = +1.65 \text{V to } +3.6 \text{V}, f_{FCLK} = 32.768 \text{kHz}, LN_BIOZ = 1, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted.}$ Typical values are at $V_{DVDD} = V_{AVDD} = +1.8 \text{V}, V_{OVDD} = +2.5 \text{V}, T_A = +25 ^{\circ}\text{C}.)$ (Note 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
PACE DETECTION	-		'			
Pace Artifact Width				0.05 to 2.0		ms
Minimum Pace Artifact Amplitude				0.5		mV
Time Resolution				16		μs
Recovery Time		Large Pacer Pulse (100mV to 700mV)		500		μs
AOUT Output Voltage Swing		f = 1kHz, THD < 0.2%		100		mV _{P-P}
INTERNAL REFERENCE/COM	MON-MODE					
V _{BG} Output Voltage	V _{BG}			0.650		V
V _{BG} Output Impedance				100		kΩ
External V _{BG} Compensation Capacitor	C _{VBG}		1			μF
V _{REF} Output Voltage	V _{REF}	T _A = +25°C	0.995	1.000	1.005	V
V _{REF} Temperature Coefficient	TC _{REF}	$T_A = 0^{\circ}C \text{ to } +70^{\circ}C$		10		ppm/°C
V _{REF} Buffer Line Regulation				330		μV/V
V _{REF} Buffer Load Regulation		$I_{LOAD} = 0$ to $100\mu A$		25		μV/μΑ
External V _{REF} Compensation Capacitor	C _{REF}		1	10		μF
V _{CM} Output Voltage	V _{CM}			0.650		V
External V _{CM} Compensation Capacitor	C _{CM}		1	10		μF
DIGITAL INPUTS (SDI, SCLK,	CSB, FCLK)					
Input-Voltage High	V _{IH}		0.7 x V _O \	/DD		V
Input-Voltage Low	V _{IL}			0.3 >	V _{OVDD}	V
Input Hysteresis	V _{HYS}		0.	.05 x V _{OVDI})	V
Input Capacitance	C _{IN}			10		pF
Input Current	I _{IN}		-1		+1	μΑ
DIGITAL OUTPUTS (SDO, INT	B, INT2B)					
Output Voltage High	V _{OH}	I _{SOURCE} = 1mA	V _{OVDD} -	0.4		V
Output Voltage Low	V _{OL}	I _{SINK} = 1mA			0.4	V
Three-State Leakage Current			-1		+1	μΑ
Three-State Output Capacitance				15		pF
POWER SUPPLY						
Analog Supply Voltage	V _{AVDD}	Connect AVDD to DVDD	1.1		2.0	V
Digital Supply Voltage	V _{DVDD}	Connect DVDD to AVDD	1.1		2.0	V
Interface Supply Voltage	V _{OVDD}	Power for I/O drivers only	1.65		3.6	V

Electrical Characteristics (continued)

 $(V_{DVDD} = V_{AVDD} = +1.1 \text{V to } +2.0 \text{V}, V_{OVDD} = +1.65 \text{V to } +3.6 \text{V}, f_{FCLK} = 32.768 \text{kHz}, LN_BIOZ = 1, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted.}$ Typical values are at $V_{DVDD} = V_{AVDD} = +1.8 \text{V}, V_{OVDD} = +2.5 \text{V}, T_A = +25 ^{\circ}\text{C}.)$ (Note 2)

PARAMETER	SYMBOL	co	NDITIONS	MIN	TYP	MAX	UNITS
			$V_{AVDD} = V_{DVDD} = +1.1V$		76		
		ECG channel	$V_{AVDD} = V_{DVDD} = +1.8V$		95		
			V _{AVDD} = V _{DVDD} = +2.0V		102	120	
		ECG channel with	$V_{AVDD} = V_{DVDD} = +1.1V$		100		
		Pace	$V_{AVDD} = V_{DVDD} = +1.8V$		124		
		(Note 2)	$V_{AVDD} = V_{DVDD} = +2.0V$		133	150	
		ECG channel with	$V_{AVDD} = V_{DVDD} = +1.1V$		114		
		Pace and AOUT	$V_{AVDD} = V_{DVDD} = +1.8V$		138		
		(Note 2)	V _{AVDD} = V _{DVDD} = +2.0V		147	190	
		ECG channel with	$V_{AVDD} = V_{DVDD} = +1.1V$		205		
		Pace, and BioZ,	$V_{AVDD} = V_{DVDD} = +1.8V$		232		
		LN_BIOZ = 0	$V_{AVDD} = V_{DVDD} = +2.0V$		242	270	
		ECG channel with	$V_{AVDD} = V_{DVDD} = +1.1V$		220		
		Pace, and BioZ,	$V_{AVDD} = V_{DVDD} = +1.8V$		247		
Supply Current	I _{AVDD} +	LN_BIOZ = 1	$V_{AVDD} = V_{DVDD} = +2.0V$		256	285	μA
		BioZ channel , LN_BIOZ = 0 (Note 2)	$V_{AVDD} = V_{DVDD} = +1.1V$		144		-
			$V_{AVDD} = V_{DVDD} = +1.8V$		163		
			$V_{AVDD} = V_{DVDD} = +2.0V$		170	190	
		BioZ channel , LN_BIOZ = 1 (Note 2)	$V_{AVDD} = V_{DVDD} = +1.1V$		158		
			$V_{AVDD} = V_{DVDD} = +1.8V$		178		
			$V_{AVDD} = V_{DVDD} = +2.0V$		185	205	-
		ECG channel and	$V_{AVDD} = V_{DVDD} = +1.1V$		186		
		BioZ, LN_BIOZ = 0	$V_{AVDD} = V_{DVDD} = +1.8V$		211		
		(Note 2)	$V_{AVDD} = V_{DVDD} = +2.0V$		220	250	
		ECG channel and	$V_{AVDD} = V_{DVDD} = +1.1V$		200		
		BioZ, LN_BIOZ = 1	$V_{AVDD} = V_{DVDD} = +1.8V$		225		
		(Note 2)	$V_{AVDD} = V_{DVDD} = +2.0V$		235	265	
		ULP Lead On	T _A = +70°C		1.3		
		Detect	T _A = +25°C		0.63	2.5	
1.1.6	la	V _{OVDD} = +1.65V, E (Note 9)	ECG channel at 512sps		0.2		
Interface Supply Current	lovdd	V _{OVDD} = 3.6V, EC (Note 9)	G channel at 512sps		0.6	1.6	μA
	I _{SAVDD} +	V _{AVDD} = V _{DVDD}	T _A = +70°C		1.3		
Shutdown Current	I _{SDVDD}	= 2.0V (Note 4)	T _A = +25°C		0.58	2.5	μA
	I _{SOVDD}	V _{OVDD} = 3.6V, V _{A\}	_{/DD} = V _{DVDD} = 2.0V			1.1	

Timing Characteristics (Note 3)

 $(V_{DVDD} = V_{AVDD} = +1.1V \text{ to } +2.0V, V_{OVDD} = +1.65V \text{ to } +3.6V, f_{FCLK} = 32.768 \text{kHz}, LN_BIOZ = 1, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } V_{DVDD} = V_{AVDD} = +1.8V, V_{OVDD} = +2.5V, T_A = +25^{\circ}C.) \text{ (Note 2)}$

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
ESD PROTECTION						
FOOD FOON DID DIN		IEC 61000-4-2 Contact Discharge (Note 10)		±8		137
ECGP, ECGN, BIP, BIN		IEC 61000-4-2 Air-Gap Discharge (Note 10)		±15		kV
All Other Pins		JEDEC JESD22-A114 HBM Transient Pulse		±2.5		kV
TIMING CHARACTERISTICS (NOTE 3)					
SCLK Frequency	f _{SCLK}		0		12	MHz
SCLK Period	t _{CP}		83			ns
SCLK Pulse Width High	t _{CH}		15			ns
SCLK Pulse Width Low	t _{CL}		15			ns
CSB Fall to SCLK Rise Setup Time	t _{CSS0}	To 1st SCLK rising edge (RE)	15			ns
CSB Fall to SCLK Rise Hold Time	t _{CSH0}	Applies to inactive RE preceding 1st RE	0			ns
CSB Rise to SCLK Rise Hold Time	t _{CSH1}	Applies to 32nd RE, executed write	10			ns
CSB Rise to SCLK Rise	t _{CSA}	Applies to 32nd RE, aborted write sequence	15			ns
SCLK Rise to CSB Fall	t _{CSF}	Applies to 32nd RE	100			ns
CSB Pulse-Width High	t _{CSPW}		20			ns
SDI-to-SCLK Rise Setup Time	t _{DS}		8			ns
SDI to SCLK Rise Hold Time	t _{DH}		8			ns
		C _{LOAD} = 20pF			40	ns
SCLK Fall to SDO Transition	t _{DOT}	C_{LOAD} = 20pF, V_{AVDD} = V_{DVDD} \geq 1.8V, V_{OVDD} \geq 2.5V			20	ns
SCLK Fall to SDO Hold	t _{DOH}	C _{LOAD} = 20pF	2			ns
CSB Fall to SDO Fall	t _{DOE}	Enable time, C _{LOAD} = 20pF			30	ns
CSB Rise to SDO Hi-Z	t _{DOZ}	Disable time			35	ns
FCLK Frequency	f _{FCLK}	External reference clock		32.768		kHz
FCLK Period	t _{FP}			30.52		μs
FCLK Pulse-Width High	t _{FH}	50% duty cycle assumed		15.26		μs
FCLK Pulse-Width Low	t _{FL}	50% duty cycle assumed		15.26		μs

- **Note 1:** All devices are 100% production tested at T_A = +25°C. Specifications over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization.
- Note 2: Guaranteed by design and characterization. Not tested in production.
- **Note 3:** One electrode drive with $<10\Omega$ source impedance, the other driven with $51k\Omega$ in parallel with a 47nF per IEC60601-2-47.
- **Note 4:** Inputs connected to $51k\Omega$ in parallel with a 47nF to V_{CM} .
- **Note 5:** Use this setting only for $V_{AVDD} = V_{DVDD} \ge 1.65V$.
- **Note 6:** Use this setting only for $V_{AVDD} = V_{DVDD} \ge 1.55V$.
- **Note 7:** Use this setting only for $V_{AVDD} = V_{DVDD} \ge 1.45V$.
- **Note 8:** This specification defines the accuracy of the calibration voltage source as applied to the ECG input, not as measured through the ADC channel.
- **Note 9:** $f_{SCLK} = 4MHz$, burst mode, EFIT = 8, $C_{SDO} = C_{INTB} = 50pF$.
- Note 10: ESD test performed with $1k\Omega$ series resistor designed to withstand 8kV surge voltage.

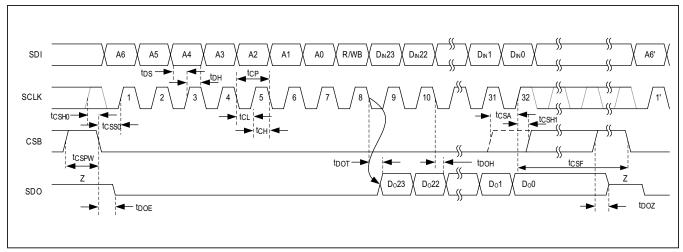


Figure 1a. SPI Timing Diagram

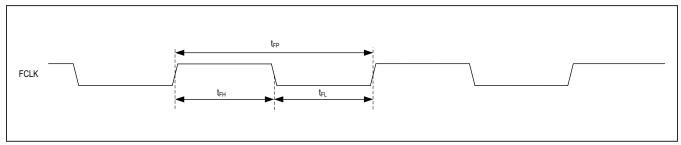
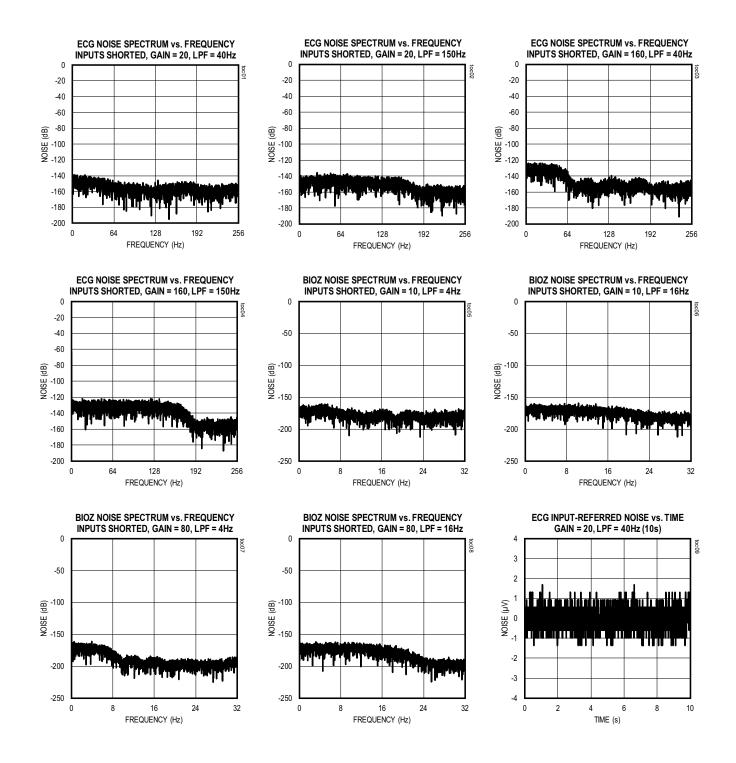


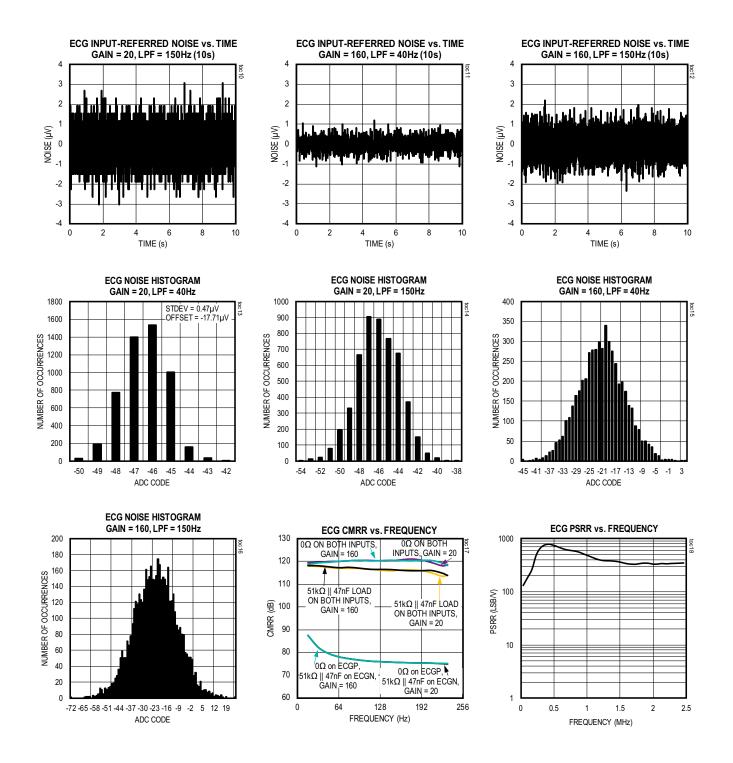
Figure 1b. FCLK Timing Diagram

Typical Operating Characteristics

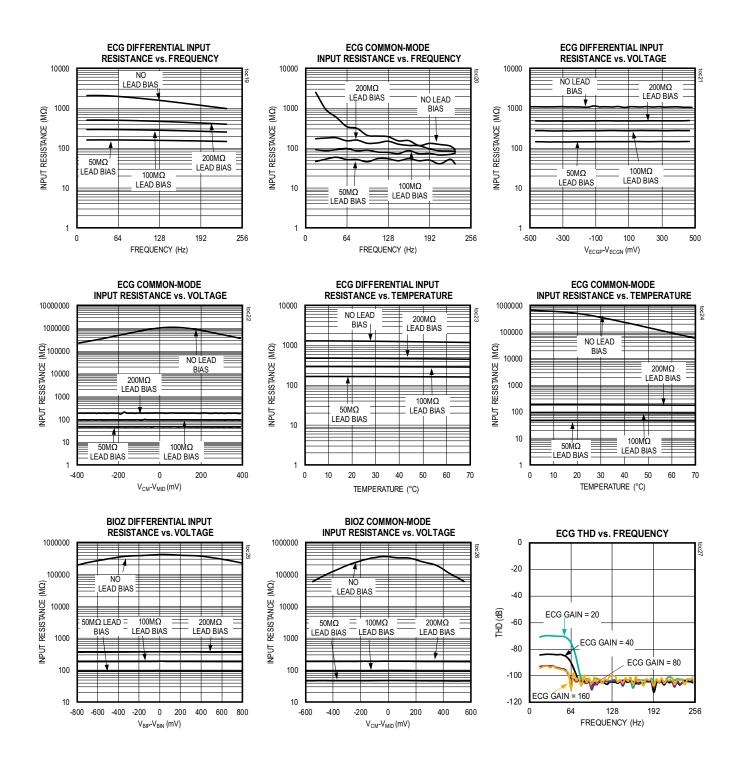
 $(V_{DVDD} = V_{AVDD} = 1.8V, V_{OVDD} = 2.5V, T_A = +25$ °C, unless otherwise noted.)



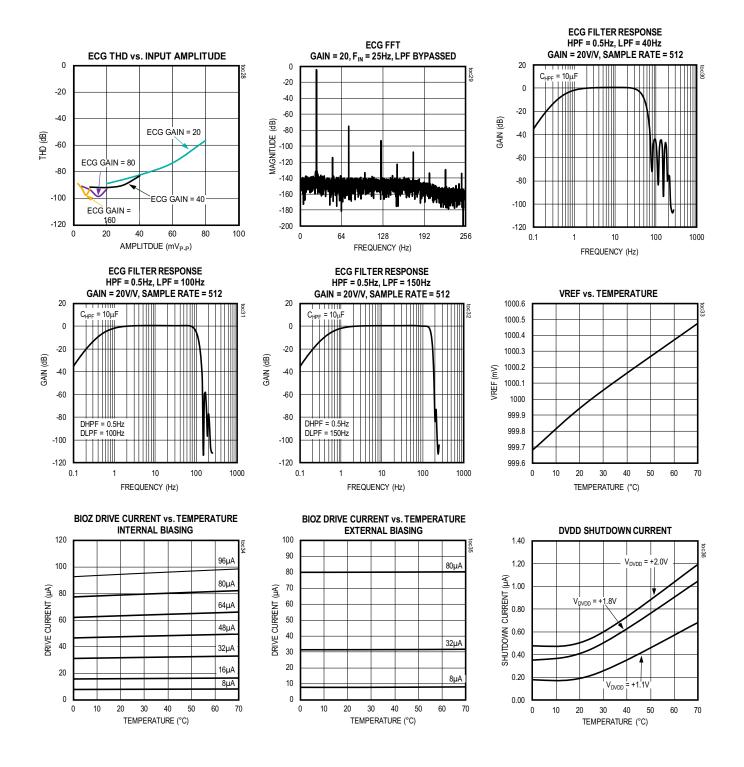
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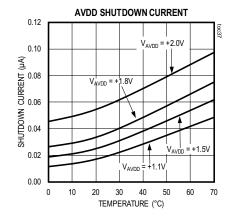
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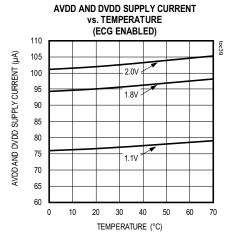


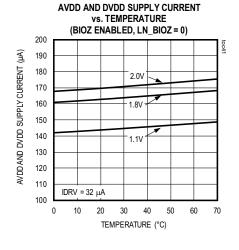
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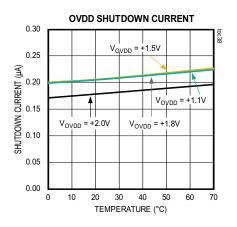


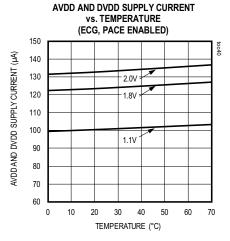
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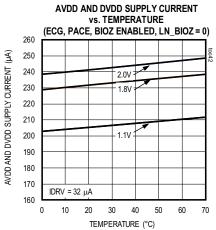






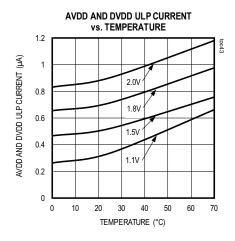


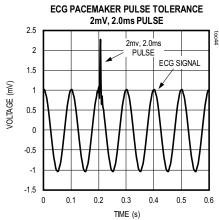


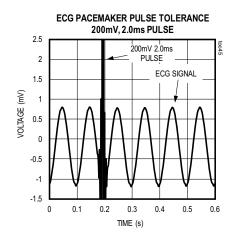


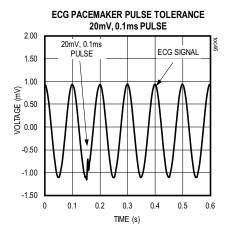
Typical Operating Characteristics (continued)

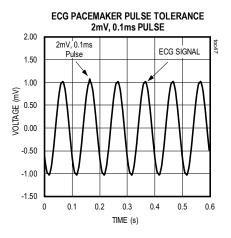
(V_{DVDD} = V_{AVDD} = 1.8V, V_{OVDD} = 2.5V, T_A = +25°C, unless otherwise noted.)



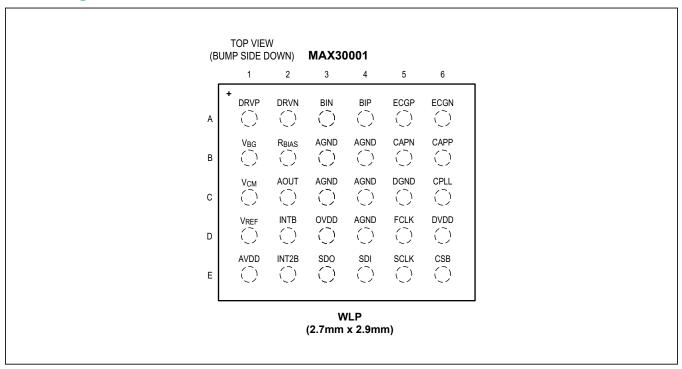








Pin Configuration



Pin Description

BUMP	NAME	FUNCTION			
WLP	NAME	FUNCTION			
A1	DRVP	Positive Output Current Source for Bio-Impedance Excitation. Requires a series capacitor between pin and electrode.			
A2	DRVN	Negative Output Current Source for Bio-Impedance Excitation. Requires a series capacitor between pin and electrode.			
A3	BIN	Bioimpedance Negative Input.			
A4	BIP	Bioimpedance Positive Input.			
A5	ECGP	ECG Positive Input.			
A6	ECGN	ECG Negative Input.			
B1	V_{BG}	Bandgap Noise Filter Output. Connect a 1.0µF X7R ceramic capacitor between V _{BG} and AGND.			
B2	R _{BIAS}	External Resistor Bias. Connect a low tempco resistor between R _{BIAS} and AGND. If external bias generator is not used then R _{BIAS} can be left floating.			
B3, B4, C3, C4, D4	AGND	Analog Power and Reference Ground. Connect into the printed circuit board ground plane.			
B5	CAPN	Analog High-Pass Filter Input. Connect a 1µF X7R capacitor (C _{HPF}) between CAPP and CAPN to form a 0.5Hz high-pass response in the ECG channel. Select a capacitor with a high voltage rating (25V) to improve linearity of the ECG signal path.			

Pin Description (continued)

BUMP		
WLP	NAME	FUNCTION
В6	CAPP	Analog High-Pass Filter Input. Connect a 1µF X7R capacitor (C _{HPF}) between CAPP and CAPN to form a 0.5Hz high-pass response in the ECG channel. Select a capacitor with a high voltage rating (25V) to improve linearity of the ECG signal path.
C1	V _{CM}	Common Mode Buffer Output. Connect a 10µF X5R ceramic capacitor between V _{CM} and AGND.
C2	AOUT	Analog Output Voltage of the Pace Channel. Programmable to select where in the signal path to output to AOUT.
C5	DGND	Digital Ground for Both Digital Core and I/O Pad Drivers. Recommended to connect to AGND plane.
C6	CPLL	PLL Loop Filter Input. Connect 1nF C0G ceramic capacitor between CPLL and AGND.
D1	V _{REF}	ADC Reference Buffer Output. Connect a 10µF X7R ceramic capacitor between V _{REF} and AGND.
D2	INTB	Interrupt Output. INTB is an active low status output. It can be used to interrupt an external device.
D3	OVDD	Logic Interface Supply Voltage.
D5	FCLK	External 32.768kHz Clock that Controls the Sampling of the Internal Sigma-Delta Converters and Decimator.
D6	DVDD	Digital Core Supply voltage. Connect to AVDD.
E1	AVDD	Analog Core Supply Voltage. Connect to DVDD.
E2	INT2B	Interrupt 2 Output. INT2B is an active-low status output. It can be used to interrupt an external device.
E3	SDO	Serial Data Output. SDO will change state on the falling edge of SCLK when CSB is low. SDO is three-stated when CSB is high.
E4	SDI	Serial Data Input. SDI is sampled into the device on the rising edge of SCLK when CSB is low.
E5	SCLK	Serial Clock Input. Clocks data in and out of the serial interface when CSB is low.
E6	CSB	Active-Low Chip-Select Input. Enables the serial interface.

Detailed Description

ECG Channel

<u>Figure 2</u> illustrates the ECG channel block diagram, excluding the ADC. The channel comprises an input MUX, a fast-recovering instrumentation amplifier, an antialias filter, and a programmable gain amplifier. The input MUX includes several features such as ESD protection, EMI filtering, lead biasing, leads off checking, and ultralow power leads-on checking. The output of this analog channel drives an 18-bit Sigma-Delta ADC.

Input MUX

The ECG input MUX shown in Figure 3 contains integrated ESD and EMI protection, DC leads off detect current sources, lead-on detect, series isolation switches, lead biasing, and a programmable calibration voltage source to enable channel built in self-test.

EMI Filtering and ESD Protection

EMI filtering of the ECGP and ECGN inputs consists of a single pole, low pass, differential, and common mode filter with the pole located at approximately 26MHz. The ECGP and ECGN inputs also have input clamps that protect the inputs from ESD events.

- ±8kV using the Contact Discharge method specified in IEC61000-4-2 ESD
- ±15kV using the Air Gap Discharge method specified in IEC61000-4-2 ESD
- For IEC61000-4-2 ESD protection, use $1k\Omega$ series resistors on ECGP and ECGN that are rated to withstand $\pm 8kV$ surge voltages.

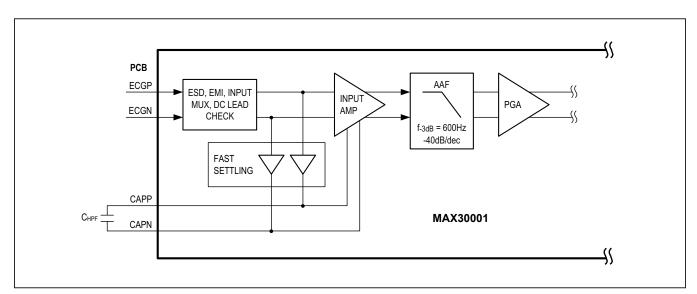


Figure 2. ECG Channel Input Amplifier and PGA Excluding the ADC

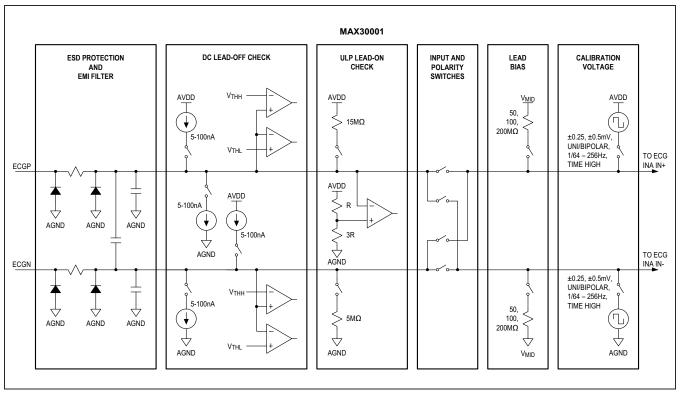


Figure 3. ECG Input MUX

DC Leads-Off Detection and ULP Leads-On Detection

The input MUX leads-off detect circuitry consists of programmable sink/source DC current sources that allow for DC leads-off detection while the channel is powered up in normal operation and an ultra-low-power (ULP) leads-on detect while the channel is powered down.

The MAX30001 accomplishes DC leads-off detection by applying a DC current to pull the ECG input voltage up to above V_{MID} + V_{TH} or down to below V_{MID} - V_{TH} . The current sources have user selectable values of 0nA, 5nA, 10nA, 20nA, 50nA, and 100nA that allow coverage of dry and wet electrode impedance ranges. Supported thresholds are V_{MID} ± 300mV (recommended), V_{MID} ±400mV, V_{MID} ± 450mV, and V_{MID} ± 500mV. A threshold of 400mV, 450mV, and 500mV must only be used when $V_{AVDD} \ge$ 1.45V, 1.55V, and 1.65V, respectively. A dynamic comparator protects against false flags generated by the input amplifier and input chopping. The comparator checks for a minimum continuous violation (or threshold exceeded) of 115ms to 140ms depending on the setting of FMSTR[1:0] before asserting any one of the LDOFF xx interrupt flags (Figure 4). See registers CNFG GEN (0x10) and CNFG EMUX (0x14) for configuration settings and see Table 1 for recommended values given electrode type and supply voltage. The 0nA setting can also be used with the V_{MID} ± 300mV threshold to monitor the input compliance of the INA when DC lead off detection is not needed.

The ULP lead on detect operates by pulling ECGN low with a pulldown resistance larger than $5M\Omega$ and pulling ECGP high with a pullup resistance larger than $15M\Omega.$ A low-power comparator determines if ECGP is pulled below a predefined threshold that occurs when both electrodes make contact with the body. When the impedance between ECGP and ECGN is less than $20M\Omega,$ an interrupt LONINT is asserted, alerting the μC to a leads-on condition.

Lead Bias

The MAX30001 limits the ECGP and ECGN DC input common mode range to V_{MID} ±150mV at V_{AVDD} = 1.1V or V_{MID} ± 550mV (typ) at V_{AVDD} = 1.8V. This range can be maintained either through external or internal lead-biasing.

Internal DC lead-biasing consists of $50M\Omega$, $100M\Omega$, or $200M\Omega$ selectable resistors to V_{MID} that drive the electrodes within the input common mode requirements of the ECG channel and can drive the connected body to the proper common mode voltage level. See register CNFG_GEN (0x10) to select a configuration.

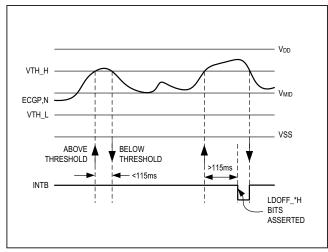


Figure 4. Lead Off Detect Behavior

The common-mode voltage, V_{CM} , can optionally be used as a body bias to drive the body to the common-mode voltage by connecting V_{CM} to a separate electrode on the body through a $200 k\Omega$ or higher resistor to limit current into the body according to IEC 60601-1:2005, 8.7.3. If this is utilized then the internal lead bias resistors to V_{MID} can be disabled.

Isolation and Polarity Switches

The series switches in the MAX30001 isolate the ECGP and ECGN pins from the internal signal path, isolating it from the subject being monitored. The series switches are disabled by default. They must be enabled to record ECG. There are also polarity switches that will swap the inputs so that ECGP goes to the minus INA input and ECGN goes to the plus INA input.

Calibration Voltage Sources

Calibration voltage sources are available to provide ± 0.25 mV (0.5mV_{P-P}) or ± 0.5 mV (1.0mV_{P-P}) inputs to the ECG channel with programmable frequency and duty cycle. The sources can be unipolar/bipolar relative to V_{MID}.

Figure 5 illustrates the possible calibration waveforms. Frequency selections are available in 4X increments from 15.625mHz to 256Hz with selected pulse widths varying from 30.5µs to 31.723ms and 50% duty cycle. Signals can be single-ended, differential, or common mode. This flexibility allows end-to-end channel-testing of the ECG signal path.

When applying calibration voltage sources with the device connected to a subject, the series input switches must be disconnected so as not to drive signals into the subject. See registers CNFG_CAL (0x12) and CNFG_EMUX (0x14) to select configuration.

Table 1. Recommended Lead Bias, Current Source Values, and Thresholds for Electrode Impedance

		ELECTRODE IMPEDANCE									
I _{DC} V _{TH}	<100kΩ	100kΩ – 200kΩ	200kΩ – 400kΩ	400kΩ – 1MΩ	1MΩ – 2MΩ	2MΩ – 4MΩ	4MΩ – 10MΩ	10MΩ – 20MΩ			
I _{DC} = 10nA		All settings of R_b $V_{TH} = V_{MID} \pm 300$ mV, ± 400 mV									
I _{DC} = 20nA	All settings of R_b All settings of V_{TH}						All settings of R _b V _{TH} =V _{MID} ± 400mV, ±450mV, ±500mV				
I _{DC} = 50nA	$\begin{array}{c} \text{All settings} \\ \text{of R}_b \\ \text{All settings of V}_{\text{TH}} \\ \text{All settings of V}_{\text{TH}} \\ \end{array} \\ \begin{array}{c} \text{All settings} \\ \text{of R}_b \\ \text{V}_{\text{TH}} = \text{V}_{\text{MID}} \\ \pm 450 \text{mV}, \\ \pm 500 \text{mV} \\ \end{array}$										
I _{DC} = 100nA	$\begin{array}{c} & \text{All settings} \\ & \text{of R}_b \\ \text{All settings of R}_b \\ \text{All settings of V}_{TH} \\ & \text{\pm400mV,} \\ & \text{\pm450mV,} \\ & \text{\pm500mV} \\ \end{array}$										

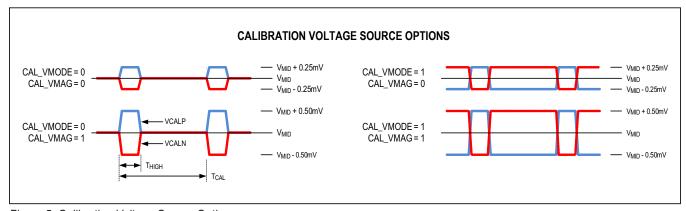


Figure 5. Calibration Voltage Source Options

Gain Settings, Input Range, and Filtering

The device's ECG channel contains an input instrumentation amplifier that provides low-noise, fixed-gain amplification (gain of 20) of the differential signal, rejects differential DC voltage due to electrode polarization, rejects common-mode interference primarily due to AC mains interference, and provides high input impedance to guarantee high CMRR even in the presence of severe electrode impedance mismatch (see Figure 2). The differential DC rejection corner frequency is set by an external capacitor (CHPF) placed between pins CAPP and CAPN, refer to Table 2 for appropriate value selection. There are three recommended options for the cutoff frequency: 5Hz, 0.5Hz, and 0.05Hz. Setting the cutoff frequency to 5Hz provides the most motion artifact rejection at the expense of ECG waveform quality, making it best suited for heart rate monitoring. For ambulatory applications requiring more robust ECG waveforms with moderate motion artifact rejection, 0.5Hz is recommended. Select 0.05Hz for patient monitoring applications in which ECG waveform quality is the primary concern and poor rejection of motion artifacts can be tolerated. The high-pass corner frequency is calculated by the following equation:

1/(2π x R_{HPF} x C_{HPF})

RHPF is specified in the Electrical Characteristics table. Following the instrumentation amplifier is a 2-pole active anti-aliasing filter with a 600Hz -3dB frequency that provides 57dB of attenuation at half the modulator sampling rate (approximately 16kHz) and a PGA with programmable gains of 1, 2, 4, and 8V/V for an overall gain of 20, 40, 80, and 160V/V. The instrumentation amplifier and PGA are chopped to minimize offset and 1/f noise. Gain settings are configured via the CNFG_ECG (0x15) register. The usable common-mode range is V_{MID} ±150mV at V_{AVDD} = 1.1V or V_{MID} ±550mV (typ) at V_{AVDD} = 1.8V. Internal lead biasing can be used to meet this requirement. The usable DC differential range is ±300mV at V_{AVDD} = 1.1V or ±650mV (typ) at V_{AVDD} = 1.8V to allow for electrode polarization voltages on each electrode. The input AC differential range is ±32.5mV or 65mV_{P-P}.

Converting ECG Samples to Voltage

ECG samples are recorded in 18-bit, left justified two's compliment format. After converting to signed magnitude format, the ECG input voltage is calculated by the following equation:

 V_{ECG} (mV) = ADC x V_{REF} / (2¹⁷ x ECG_GAIN)

ADC is the ADC counts in signed magnitude format, V_{REF} is 1000mV (typ) (refer to the <u>Electrical Characteristics</u> section), and ECG_GAIN is 20V/V, 40V/V, 80V/V, or 160V/V, set in CNFG_ECG (0x15).

Fast Recovery Mode

The input instrumentation amplifier has the ability to rapidly recover from an excessive overdrive event such as a defibrillation pulse, high-voltage external pacing, and electro-surgery interference. There are two modes of recovery that can be used: automatic or manual recovery. The mode is programmed by the FAST[1:0] bits in the MNGR_DYN (0x05) register.

Table 2. ECG Analog HPF Corner Frequency Selection

C _{HPF}	HPF CORNER FREQUENCY
0.1µF	≤ 5Hz
1.0µF	≤ 0.5Hz
10μF	≤ 0.05Hz

Table 3. Fast Recovery Mode Recovery Time vs. Number of Samples

SAMPLE RATE (sps)	NUMBER OF SAMPLES	RECOVERY TIME (APPROXIMATE) (ms)
512	255	498
256	127	496
128	63	492
500	249	498
250	124	496
125	64	512
200	99	495
199.8	99	495.5

MAX30001

Ultra-Low-Power, Single-Channel Integrated Biopotential (ECG, R-to-R, and Pace Detection) and Bioimpedance (BioZ) AFE

Automatic mode engages once the saturation counter exceeds approximately 125ms (t_{SAT}). The counter is activated the first time the ADC output exceeds the symmetrical threshold defined by the FAST_TH[5:0] bits in the MNGR_DYN (0x05) register and accumulates the time that the ADC output exceeds either the positive or negative threshold. If the saturation counter exceeds 125ms, it triggers the fast settling mode (if enabled) and resets. The saturation counter can also be reset prior to triggering the fast settling mode if the ADC output falls below the threshold continuously for 125ms (t_{BLW}). This feature is designed to avoid false triggers due to the QRS complex. Once triggered, fast settling mode is engaged for 500ms.

(<u>Table 3</u>). ECG samples are tagged if they were taken while fast settling mode was asserted (Figure 6).

In manual mode, a user algorithm running on the host microcontroller or an external stimulus input will generate the trigger to enter fast recovery mode. The host microcontroller then enables the manual fast recovery mode in the MNGR_DYN (0x05) register. The manual fast recovery mode can be of a much shorter duration than the automatic mode and allows for more rapid recovery. One such example is recovery from external high-voltage pacing signals in a few milliseconds to allow the observation of a subsequent p-wave.

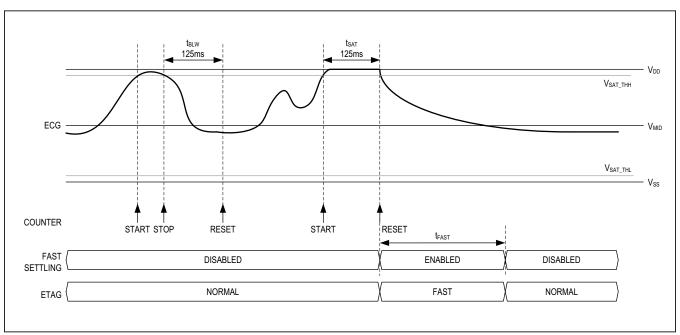


Figure 6. Automatic Fast Settling Behavior

Decimation Filter

The decimation filter consists of an FIR decimation filter to the data rate followed by a programmable IIR and FIR filter to implement HPF and LPF selections.

The high-pass filter options include a 1st-order IIR Butterworth filter with a 0.4Hz corner frequency along with a pass through setting for DC coupling. Low-pass filter options include a 12-tap linear phase (constant group

delaya approximately 40Hz, 100Hz, or 150Hz corner frequencies, depending on the sampling rate. See register CNFG_ ECG (0x15) to configure the filters. <u>Table 4</u> illustrates the ECG latency in samples and time for each ADC data rate.

Noise Measurements

<u>Table 5</u> shows the noise performance of the ECG channel of MAX30001 referred to the ECG inputs.

Table 4. ECG Latency in Samples and Time as a Function of ECG Data Rate and Decimation

ECG	CHANNEL SETTIN	IGS	LATENCY				
INPUT SAMPLE RATE (Hz)	OUTPUT DATA RATE (sps)	DECIMATION RATIO	WITHOUT LPF (INPUT SAMPLES)	WITH LPF (INPUT SAMPLES)	WITHOUT LPF (ms)	WITH LPF (ms)	
32,768	512	64	650	1,034	19.836	31.555	
32,000	500	64	650	1,034	20.313	32.313	
32,768	256	128	2,922	3,690	89.172	112.610	
32,000	250	128	2,922	3,690	91.313	115.313	
32,000	200	160	1,242	2,202	38.813	68.813	
31,968	199.8	160	1,242	2,202	38.851	68.881	
32,768	128	256	3,370	4,906	102.844	149.719	
32,000	125	256	3,370	4,906	105.313	153.313	

Table 5. ECG Channel Noise Performance

GAIN	BANDWIDTH	NC	DISE	SNR	ENOB
V/V	Hz	μV _{RMS}	μV _{P-P}	dB	Bits
	40	0.46	3.04	97.7	15.9
20	100	0.64	4.20	94.9	15.5
	150	0.77	4.60	93.2	15.2
	40	0.40	2.64	92.9	15.1
40	100	0.54	3.56	90.3	14.7
	150	0.66	4.34	88.6	14.4
	40	0.35	2.31	88.0	14.3
80	100	0.50	3.33	84.9	13.8
	150	0.62	4.09	83.1	13.5
	40	0.34	2.22	82.4	13.4
160	100	0.49	3.24	79.1	12.8
	150	0.61	4.01	77.2	12.5

R-to-R Detection

The MAX30001 contains built-in hardware to detect R-R intervals using an adaptation of the Pan-Tompkins QRS detection algorithm*. The timing resolution of the R-R interval is approximately 8ms and depends on the setting of FMSTR [1:0] in CNFG_GEN (0x10) register. See Table 26 for the timing resolution of each setting.

When an R event is identified, the RRINT status bit is asserted and the RTOR_REG (0x25) register is updated with the count seen since the last R event. Figure 7 illustrates the R-R interval on a QRS complex. Refer to registers CNFG_RTOR1 (0x1D) and CNFG_RTOR2 (0x1E) for selection details.

The latency of the R-to-R value written to the RTOR Interval Memory Register is the sum of the R-to-R decimation delay and the R-to-R detection delay blocks. The R-to-R decimation factor is fixed at 256 and the decimation delay (t_{R2R_DEC}) is always 3,370 FMSTR clocks, as shown in Table $\overline{6}$.

The detection circuit consists of several digital filters and signal processing delays. These depend on the WNDW[3:0] bits in the CNFG_RTOR (0x1D) register. The detection delay (t_{R2R_DET}) is described by the following equation:

t_{R2R_DET} = 5,376 + 256 x WNDW in FMSTR clocks where WNDW is an integer from 0 to 15

and the total latency (t_{R2R_DEL}) is the sum of the two delays and summarized in the equation below:

 $t_{R2R_DEL} = t_{R2R_DEC} + t_{R2R_DET} = 3,370 + 5,376 + 256 \times WNDW$ in FMSTR clocks where WNDW is an integer from 0 to 15.

The total R-to-R latency minus the ECG latency is the delay of the R-to-R value relative to the ECG data and can be used to place the first R-to-R value on the ECG data plot. The succeeding values in the R-to-R Interval Memory Register can be used as is to locate subsequent R-to-R values on the ECG data plot relative to the initial placement.

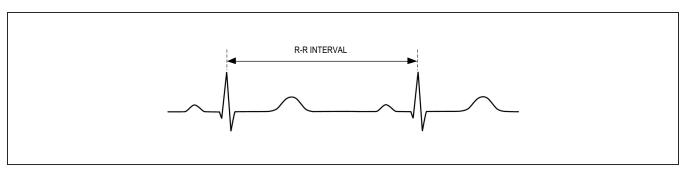


Figure 7. R-to-R Interval Illustration

Table 6. R-to-R Decimation Delay vs. Register Settings

EMCTD [4:0]	EMETE EDEO	FMSTR FREQ		RTOR TIME RESOLUTION	DELAY IN R-TO-	-R DECIMATION	
FMSTR [1:0]	FWSTRFREQ	(Hz)	DECIMATION	(ms)	FMSTR CLKs (ms)		
00	FCLK	32,768	256	7.8125	3370	102.844	
01	FCLK x 625/640	32,000	256	8.0	3370	105.313	
10	FCLK x 625/640	32,000	256	8.0	3370	105.313	
11	FCLK x 640/656	31,968.78	256	8.0078	3370	105.415	

^{*}J. Pan and W.J. Tompkins, "A Real-Time QRS Detection Algorithm," IEEE Trans. Biomed. Eng., vol. 32, pp. 230-236

Pace Channel

MAX30001 provides an analog based pace detection for up to three chamber pacing with data logging and ECG tagging for up to three rising and falling edges per ECG sample. See register CNFG_PACE (0x1A) to select configuration and ECG FIFO and PACE memory for detailed descriptions of the ECG and PACE FIFOs.

Real time monitoring of pace edge events can be accomplished by unmasking PEDGE via EN_INT (0x02) and EN_INT2 (0x03) and using the self-clear behavior; see CLR PEDGE=1 in register MNGR INT (0x04).

Current injection rates for Bio-Impedance measurements are limited to 40kHz and 80kHz when pace detection is enabled to avoid glitches caused by current injection

being interpreted as a pace event. A single-ended analog signal is provided at pin AOUT to allow digitization of the PACE pulses with an external analog to digital converter. See register CNFG_PACE (0x1A) for gain, low pass and high pass filter options and AOUT signal selection.

BioZ Channel

<u>Figure 8</u> illustrates the BioZ channel block diagram, excluding the ADC. The channel comprises an input MUX, an instrumentation amplifier, a mixer, an anti-alias filter, and a programmable gain amplifier. The MUX includes several features such as ESD protection, EMI filtering, lead biasing, leads off checking, and ultra-low power leads-on checking. The output of this analog channel drives a 20-bit Sigma-Delta ADC.

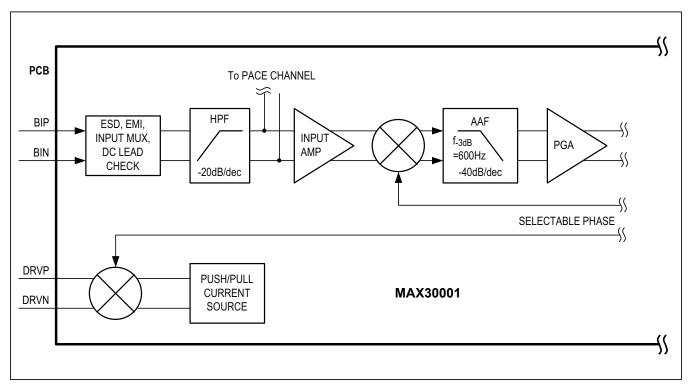


Figure 8. BioZ Channel Input Amplifier, Mixer, and PGA Excluding the ADC and Current Drive Output

Input MUX

The BioZ input MUX shown in Figure 9 contains integrated ESD and EMI protection, DC leads off detect current sources and comparators, lead-on detect, series isolation switches, lead biasing, a programmable calibration voltage source to enable channel built in self-test for the pace channel, and a built in programmable resistor load.

EMI Filtering and ESD Protection

EMI filtering of the BIP and BIN inputs consists of a single pole, low pass, differential, and common mode filter with

the pole located at approximately 26MHz. The BIP and BIN inputs also have input clamps that protect the inputs from ESD events.

- ±8kV using the Contact Discharge method specified in IEC61000-4-2 ESD
- ±15kV using the Air Gap Discharge method specified in IEC61000-4-2 ESD
- For IEC61000-4-2 ESD protection, use 1kΩ series resistors on BIP and BIN that is rated to withstand ±8kV surge voltages

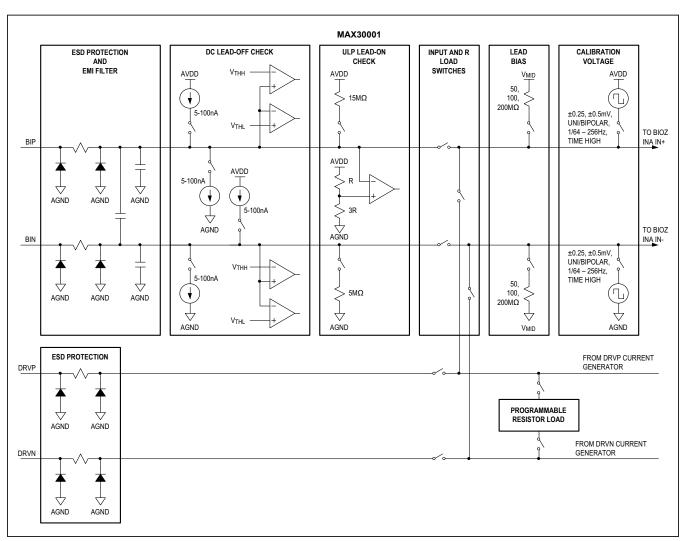


Figure 9. BioZ Input MUX

Leads-Off Detection and ULP Leads-On Detection

MAX30001 provides the capability of detecting lead off scenarios that involve two electrode and four electrode configurations through the use of digital threshold and analog threshold comparisons. There are three methods to detect lead-off for the BioZ channel. There is a compliance monitor for the current generator on the DRVP and DRVN pins detecting when the voltage on the pins is outside its operating range. The BIOZ_CGMON bit in the CNFG BIOZ (0x18) register enables this function and the BCGMON, BCGMP, and BCGMN bits in the STATUS (0x01) register indicate if the DRVP and DRVN pins are out of compliance. There is a DC lead-off circuit on the BIP and BIN pins (same as on the ECGP and ECGN pins, see ECG description) that sinks or sources a programmable DC current and window comparators with a programmable threshold to detect the condition. There is a digital AC lead off detection monitoring the output of the BioZ ADC with programmable under and overvoltage levels performing a digital comparison. The EN BLOFF bit in the CNFG GEN (0x10) register enables this function and the BLOFF_HI_IT[7:0] and BLOFF_LO_ IT[7:0] bits in the MNGR_DYN (0x05) register sets the digital threshold for detection. Refer to Table 7 for lead off conditions and register settings to allow detection. The 0nA setting can also be used with the $V_{\mbox{\scriptsize MID}}$ ± 300mV threshold to monitor the input compliance of the INA when DC lead off detection is not needed.

The ULP lead-on detect operates by pulling BIN low with a pulldown resistance larger than $5M\Omega$ and pulling BIP high with a pullup resistance larger than $15M\Omega.$ A low-power comparator determines if BIP is pulled below a predefined threshold that occurs when both electrodes make contact with the body. When the impedance between BIP and BIN is less than $20M\Omega,$ an interrupt LONINT is asserted, alerting the μC to a leads-on condition.

Lead Bias

The MAX30001 limits the BIP and BIN DC input common mode range to V_{MID} ±150mV at V_{AVDD} = 1.1V or V_{MID} ±550mV (typ) at V_{AVDD} = 1.8V. This range can be maintained either through external/internal lead-biasing.

Internal DC lead-biasing consists of $50M\Omega$, $100M\Omega$, or $200M\Omega$ selectable resistors to V_{MID} that drive the electrodes within the input common mode requirements of the ECG channel and can drive the connected body to the proper common mode voltage level. See the EN_RBIAS[1:0], RBIASV[1:0], RBIASP, and RBIASN bits in the CNFG GEN (0x10) register to select a configuration.

Table 7. BioZ Lead Off Detection Configurations

CONFIGURATION	CONDITION	DRVP/N	BIP/N	MEASURED SIGNAL	REGISTER SETTING TO DETECT
Two-Electrode (Shared DRV/BI)	1 Electrode Off	Rail to Rail	Rail to Rail	Rail to Rail (Saturated Inputs)	CNFG_GEN (0x10), EN_BLOFF[1:0] = 10 or 11 MNGR_DYN (0x05), BLOFF_HI_IT[7:0]
	1 DRV Electrode Off, Large Body Coupling	Rail to Rail	Normal	½ Signal	CNFG_BIOZ (0x18), BIOZ_CGMON=1
Four-Electrode	1 DRV Electrode Off, Small Body Coupling	Rail to Rail	Rail to Rail	Rail to Rail (Saturated Inputs)	CNFG_GEN (0x10), EN_BLOFF[1:0] = 10 or 11 MNGR_DYN (0x05), BLOFF_HI_IT[7:0]
(Force/Sense)	1 BI (sense) Electrode Off	Normal	Floating	½ Signal	CNFG_GEN (0x10), EN_DCLOFF=10
	Both BIP/N (sense) Electrodes Off	Normal	Floating	No Signal	CNFG_GEN (0x10), EN_BLOFF[1:0] = 01 or 11 MNGR_DYN (0x05), BLOFF_LO_IT[7:0]
	1 DRV and 1 BI Electrode Off	Rail to Rail	Wide Swing, Dependent on Body Coupling	Rail to Rail	CNFG_GEN (0x10), EN_BLOFF[1:0] = 10 or 11 MNGR_DYN (0x05), BLOFF_HI_IT[7:0]

MAX30001

Ultra-Low-Power, Single-Channel Integrated Biopotential (ECG, R-to-R, and Pace Detection) and Bioimpedance (BioZ) AFE

The common-mode voltage, V_{CM} , can optionally be used as a body bias to drive the body to the common-mode voltage by connecting V_{CM} to a separate electrode on the body through a 200k Ω or higher resistor to limit current into the body according to IEC 60601-1:2005, 8.7.3. If this is utilized then the internal lead bias resistors to V_{MID} can be disabled. If ECGP/ECGN pins are shared with the BIP/BIN pins then it is only necessary to enable lead bias on ECG or BioZ.

Calibration Voltage Sources

Calibration voltage sources are available to provide ± 0.25 mV (0.5mV $_{P-P})$ or ± 0.5 mV (1.0mV $_{P-P})$ inputs to the BioZ/Pace channel with programmable frequency and duty cycle. The sources can be unipolar/bipolar relative to V $_{MID}$.

 $\underline{\text{Figure }10}$ illustrates the possible calibration waveforms. Frequency selections are available in 4X increments from 15.625mHz to 256Hz with selected pulse widths varying from 30.5µs to 31.723ms and 50% duty cycle. Signals can be single-ended, differential, or common mode. This flexibility allows end-to-end channel-testing of the Pace signal path and is primarily used for pacemaker pulse detection validation.

When applying calibration voltage sources with the device connected to a subject, the series input switches must be disconnected so as not to drive signals into the subject. See registers CNFG_CAL (0x12) and CNFG_BMUX (0x14) to select configuration.

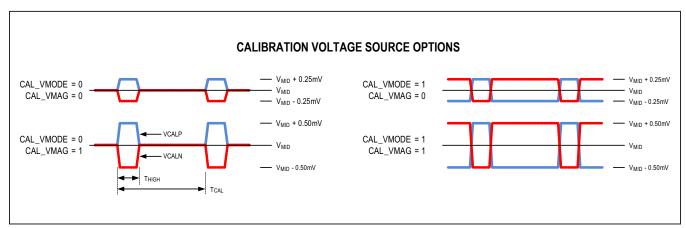


Figure 10. Calibration Voltage Source Options

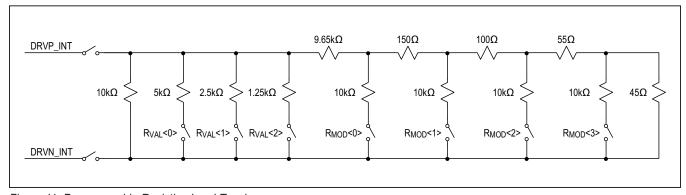


Figure 11. Programmable Resistive Load Topology

Programmable Resistive Load

The programmable resistive load on the DRVP/DRVN pins allows a built in self-test of the current generator (CG) and the entire BioZ channel. Refer to Figure 11 for implementation details.

Nominal resistance can be varied between $5k\Omega$ and 625Ω . The modulation resistance is used to switch the load resistance between R_{NOM} and $(R_{NOM} - R_{MOD})$ at the selected modulation rate. The modulation resistance is dependent on the nominal resistance value with resolution of $50.4m\Omega$ to 2.96Ω at the largest nominal resistance $(5k\Omega)$ and $15.3m\Omega$ to $46.3m\Omega$ with the smallest nominal resistance (625Ω) . Refer to Table 8 for a complete listing

of nominal and modulated resistor values. Modulation rate can be programmed between 625mHz to 4Hz.

See registers CNFG_CAL (0x12) and CNFG_BMUX (0x17) to select configuration for modulation rate and resistor value.

Current Generator

The current generator provides square-wave modulating differential current that is AC injected into the body via pins DRVP and DRVN with the bio-impedance sensed differentially through pins BIP and BIN. Two and four electrode configurations are supported for typical wet and dry electrode impedances.

Table 8. Programmable Resistive Load Values

R _{NOM} (Ω)	R _{MOD}	R _{VAL}			R _{MOD}			
NOW (/	(mΩ)	<2>	<1>	<0>	<3>	<2>	<1>	<0>
	-	0	0	0	0	0	0	0
5000.000	2960.7	0	0	0	0	0	0	1
5000.000	980.6	0	0	0	0	0	1	0
	247.5	0	0	0	0	1	0	0
	-	0	0	1	0	0	0	0
2500.000	740.4	0	0	1	0	0	0	1
2500.000	245.2	0	0	1	0	0	1	0
	61.9	0	0	1	0	1	0	0
	-	0	1	0	0	0	0	0
1666.667	329.1	0	1	0	0	0	0	1
1000.007	109.0	0	1	0	0	0	1	0
	27.5	0	1	0	0	1	0	0
	-	0	1	1	0	0	0	0
1250.000	185.1	0	1	1	0	0	0	1
	61.3	0	1	1	0	0	1	0
	-	1	0	0	0	0	0	0
1000.000	118.5	1	0	0	0	0	0	1
	39.2	1	0	0	0	0	1	0
	-	1	0	1	0	0	0	0
833.333	82.3	1	0	1	0	0	0	1
	27.2	1	0	1	0	0	1	0
	-	1	1	0	0	0	0	0
714.286	60.5	1	1	0	0	0	0	1
	20.0	1	1	0	0	0	1	0
	-	1	1	1	0	0	0	0
625.000	46.3	1	1	1	0	0	0	1
	15.3	1	1	1	0	0	1	0

Current amplitudes between $8\mu A_{PK}$ to $96\mu A_{PK}$ are selectable with current injection frequencies between 125Hz and 131.072kHz in power of two increments. See register CNFG_BIOZ (0x18) for configuration selections.

Current amplitude should be chosen so as not exceed $90\text{mV}_{\text{P-P}}$ at the BIP and BIN pins based on the network impedance at the current injection frequency. A 47nF DC blocking capacitor is required between both DRVP and DRVN and their respective electrodes.

The current generator also includes a phase offset adjustment, which delays the drive current modulator with respect to the input mixer. The phase can be adjusted in 11.25° increments from 0° to 168.75° for injection frequencies up to f_{MSTR}. For injection frequencies of 2 x f_{MSTR} and 4 x f_{MSTR}, the phase resolution is reduced to 22.5° and 45° respectively. See CNFG_BIOZ (0x18) for details.

Converting BioZ Samples to Ωs

BioZ samples are recorded in 20-bit, left justified two's compliment format. After converting to signed magnitude format, BioZ is calculated by the following equation:

BioZ (Ω) = ADC x
$$V_{REF}$$
 / (2¹⁹ x BIOZ_CGMAG x BIOZ_GAIN)

ADC is the ADC counts in signed magnitude format, V_{REF} is 1V (typ) (refer to the <u>Electrical Characteristics</u> section), BIOZ_CGMAG is 8 to 96 x 10⁻⁶A, and BIOZ_GAIN is 10V/V, 20V/V, 40V/V, or 80V/V. BIOZ_CGMAG and BIOZ_GAIN are set in CNFG_BIOZ (0x18).

Current Selection and Resolution Calculation Example 1 (Two Terminal with Common Protection)

Selection of the appropriate current is accomplished by first calculating the resistive component of the network impedance at the injection frequency. Worst case electrode impedances should be used.

Given <u>Figure 12</u> and a current injection frequency of 80kHz, the resistive component of the network impedance is:

$$R_{BODY} + 2R_{P1} + 2R_{P2} + 2R_{S} + Re\{\frac{2R_{E}}{1 + j\omega R_{E}C_{E}}\} = 2.7k\Omega$$

where R_{BODY} = 100 Ω , R_{P1} = 1k Ω , R_{P2} = 200 Ω , R_{S} = 100 Ω , R_{E} = 1M Ω , C_{E} = 5nF. The maximum current injection is the maximum AC input differential range (90mV_{PK}) divided by the network impedance (2.7k Ω) or 33.3µA_{PK}. The closest selectable lower value is 32µA_{PK}.

Given the current injection value and the channel bandwidth (refer to register CNFG_BIOZ (0x18) for digital LPF selection) the resolvable impedance can be calculated by dividing the appropriate input referred noise by the current injection value. For example, with a bandwidth of 4Hz, the input referred noise with a gain of 20V/V is $0.16\mu V_{RMS}$ or $1.1\mu V_{P-P}$. The resolvable impedance is therefore $1.1\mu V_{P-P}$ / $32\mu A_{PK}$ = $69m\Omega_{P-P}$ or $5m\Omega_{RMS}$.

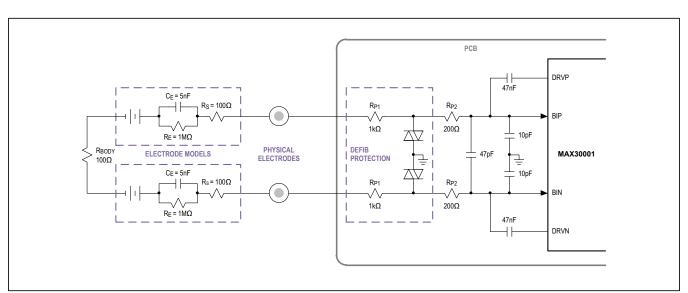


Figure 12. Example Configuration – Two Terminal with Common Protection

Current Selection and Resolution Calculation Example 2 (Four Terminal)

Selection of the appropriate current is accomplished by first calculating the resistive component of the network impedance at the injection frequency. Worst case electrode impedances should be used.

Given Figure 13 and a current injection frequency of 80kHz, the resistive component of the network impedance is:

$$R_{BODY} + 2R_{DP1} + 2R_{DP2} + 2R_S + Re\left\{\frac{2R_E}{1 + j\omega R_E C_E}\right\} = 2.7k\Omega$$

where R_{BODY} = 100 $\Omega,~R_{DP1}$ = 1k $\Omega,~R_{DP2}$ = 200 $\Omega,~R_{S}$ = 100 $\Omega,~R_{E}$ = 1M $\Omega,~C_{E}$ = 5nF. The maximum current injection is the maximum DRVP/N Compliance Voltage (V_{DD}-0.5V = 0.6V for V_{DD} = 1.1V) divided by the network impedance (2.7k Ω) or 222.2µA_{PK}. The closest selectable lower value is $96\mu A_{PK}$.

Given the current injection value and the channel bandwidth (refer to register CNFG_BIOZ (0x18) for digital LPF

selection) the resolvable impedance can be calculated by dividing the appropriate input referred noise by the current injection value. For example, with a bandwidth of 4Hz, the input referred noise with a gain of 40V/V is $0.12\mu V_{RMS}$ or $0.78\mu V_{P-P}$. The resolvable impedance is therefore $0.78\mu V_{P-P}/96\mu A_{PK}=8m\Omega_{P-P}$ or $1.2m\Omega_{RMS}$.

Decimation Filter

The decimation filter consists of an FIR decimation filter to the data rate followed by a programmable IIR and FIR filter to implement HPF and LPF selections.

The high-pass filter options include a fourth-order IIR Butterworth filter with a 0.05Hz or 0.5Hz corner frequency along with a pass through setting for DC coupling. Lowpass filter options include a 12-tap linear phase (constant group delay) FIR filter with 4Hz, 8Hz, or 16Hz corner frequencies. See register CNFG_BIOZ (0x18) to configure the filters. Table 9 illustrates the BioZ latency in samples and time for each ADC data rate.

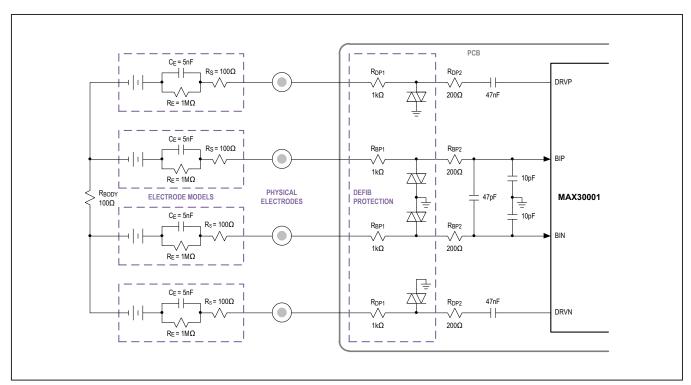


Figure 13. Example Configuration—Four Terminal

Noise Measurements

<u>Table 10</u> shows the noise performance of the BioZ channel of MAX30001 referred to the BioZ inputs.

Reference and Common Mode Buffer

The MAX30001 features internally generated reference voltages. The bandgap output (V_{BG}) pin requires an external 1.0µF capacitor to AGND and the reference output (V_{REF}) pin requires a 10µF external capacitor to AGND for compensation and noise filtering.

A common-mode buffer is provided to buffer 650mV which is used to drive common mode voltages for internal blocks. Use a 10µF external capacitor between V_{CM} to AGND to provide compensation and noise filtering. The common-mode voltage, V_{CM} , can optionally be used as a body bias to drive the body to the common-mode voltage by connecting V_{CM} to a separate electrode on the body through a 200k Ω or higher resistor to limit current into the body according to IEC 60601-1:2005, 8.7.3. If this is utilized then the internal lead bias resistors to V_{MID} may be disabled if the input signals are within the common-mode input range.

Table 9. BioZ Latency in Samples and Time as a Function of BioZ Data Rate and Decimation

BioZ	CHANNEL SETTI	NGS	LATENCY				
INPUT SAMPLE RATE (Hz)	OUTPUT DATA RATE (sps)	DECIMATION RATIO	WITHOUT LPF (INPUT SAMPLES)	WITH LPF (INPUT SAMPLES)	WITHOUT LPF(ms)	WITH LPF (ms)	
32,768	64	512	3,397	6,469	103.668	197.418	
32,000	62.5	512	3,397	6,469	106.156	202.156	
32,000	50	640	5,189	9,029	162.156	282.156	
31,968	49.95	640	5,189	9,029	162.319	282.439	
32,768	32	1,024	7,557	13,701	230.621	418.121	
32,000	31.25	1,024	7,557	13,701	236.156	428.156	
32,000	25	1,280	9,605	17,285	300.156	540.156	
31,968	24.975	1,280	9,605	17,285	300.457	540.697	

Table 10. BioZ Channel Noise Performance

GAIN	BANDWIDTH	NO	OISE	SNR	ENOB
V/V	Hz	μV _{RMS}	μV _{P-P}	dB	Bits
	4	0.23	1.55	101.6	16.6
10	8	0.28	1.87	100.0	16.3
	16	0.35	2.34	98.0	16.0
	4	0.16	1.10	104.9	17.1
20	8	0.19	1.27	103.4	16.9
	16	0.26	1.68	100.9	16.5
	4	0.12	0.78	107.6	17.6
40	8	0.16	1.07	104.9	17.1
	16	0.22	1.48	102.0	16.7
	4	0.11	0.72	108.3	17.7
80	8	0.15	1.01	105.3	17.2
	16	0.21	1.42	102.4	16.7

 $SNR = 20log(V_{IN}(RMS)/V_{N}(RMS)), ENOB = (SNR - 1.76)/6.02$

 $V_{IN(P-P)}$ = 100mV, V_{INRMS} = 35.4mV for a gain of 10V/V. The input amplitude is reduced accordingly for high gain settings.

SPI Interface Description

32 Bit Normal Mode Read/Write Sequences

The MAX30001 interface is SPI/QSPI/Micro-wire/DSP compatible. The operation of the SPI interface is shown in Figure 1a. Data is strobed into the MAX30001 on SCLK rising edges. The device is programmed and accessed by a 32 cycle SPI instruction framed by a CSB low interval. The content of the SPI operation consists of a one byte command word (comprised of a seven bit address and a Read/Write mode indicator, i.e., A[6:0] + R/W) followed by a three-byte data word. The MAX30001 is compatible with CPOL = 0/CPHA = 0 and CPOL = 1/CPHA = 1 modes of operation.

Write mode operations will be executed on the 32nd SCLK rising edge using the first four bytes of data available. In write mode, any data supplied after the 32nd SCLK rising edge will be ignored. Subsequent writes require CSB to de-assert high and then assert low for the next write command. In order to abort a command sequence, the rise of CSB must precede the updating (32nd) rising-edge of SCLK, meeting the $t_{\rm CSA}$ requirement.

Read mode operations will access the requested data on the 8th SCLK rising edge, and present the MSB of the requested data on the following SCLK falling edge, allowing the μ C to sample the data MSB on the 9th SCLK rising edge. Configuration, Status, and FIFO data are all available via normal mode read back sequences. If more than 32 SCLK rising edges are provided in a normal read sequence then the excess edges will be ignored and the device will read back zeros.

If accessing the STATUS register or the ECG, BioZ or PACE FIFO memories, all interrupt updates will be made

and the internal FIFO read pointer will be incremented in response to the 30th SCLK rising edge, allowing for internal synchronization operations to occur. See the data tag structures used within each FIFO for means of detecting end-of-file (EOF) samples, invalid (empty samples) and other aides for efficiently using and managing normal mode read back operations.

Burst Mode Read Sequence

The MAX30001 provides commands to read back the ECG, BioZ or PACE FIFO memory in a burst mode to increase data transfer efficiency. Burst mode uses different register addresses than the normal read sequence register addresses. A modified burst mode is supported for each PACE FIFO word group (see description of PACE0 to PACE5 register group). The first 32 SCLK cycles operate exactly as described for the normal mode. If the µC continues to provide SCLK edges beyond the 32nd rising edge, the MSB of the next available FIFO word will be presented on the next falling SCLK edge, allowing the µC to sample the MSB of the next word on the 33rd SCLK rising edge. Any affected interrupts and/or FIFO read pointers will be incremented in response to the (30+nx24)th SCLK rising edge where n is an integer starting at 0. (i.e., on the 30th, 54th, and 78th SCLK risingedges for a three-word, burst-mode transfer).

This mode of operation will continue for every 24 cycle sub frame, as long as there is valid data in the FIFO. See the data tag structures used within each FIFO for means of detecting end-of-file (EOF) samples, invalid (empty samples) and other aides for efficiently using and managing burst mode read back operations.

There is no burst mode equivalent in write mode.

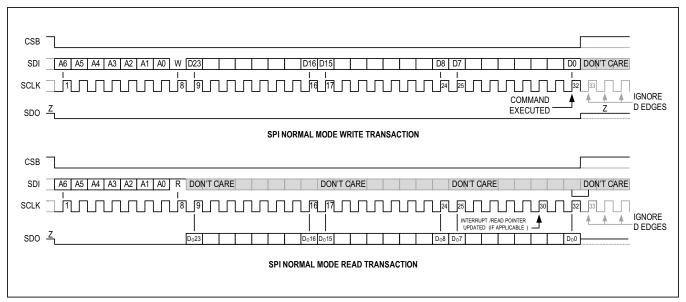


Figure 14. SPI Normal Mode Transaction Diagram

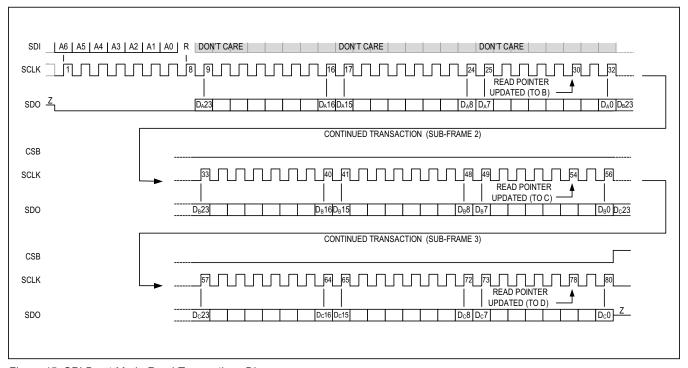


Figure 15. SPI Burst Mode Read Transactions Diagram

User Command and Register Map

REG		R/W				DA	ATA INDEX						
[6:0]	NAME	MODE	23/15/7	22/14/6	21/13/5	20/12/4	19/11/3	18/10/2	17/9/1	16/8/0			
0x00	NO-OP	R/W	x / x / x	x/x/x	x/x/x	x/x/x	x/x/x	x / x / x	x/x/x	x/x/x			
			EINT	EOVF	FSTINT	DCLO FFINT	BINT	BOVF	BOVER	BUNDR			
0x01	STATUS	R	BCGMON	PINT	POVF	PEDGE	LONINT	RRINT	SAMP	PLLINT			
			Х	х	BCGMP	BCGMN	LDOFF_PH	LDOFF_PL	LDOFF_NH	LDOFF_NL			
0x02	EN_INT	544	EN_EINT	EN_EOVF	EN_ FSTINT	EN_ DCLOFFINT	EN_BINT	EN_BOVF	EN_BOVER	EN_BUNDR			
0x03	EN_INT2	R/W	EN_BCGMON	EN_PINT	EN_POVF	EN_PEDGE	EN_LONINT	EN_ RRINT	EN_SAMP	EN_ PLLINT			
			Х	х	х	х	х	Х	INTB_T	YPE[1:0]			
					EFIT[4:0]				BFIT[2:0]				
0x04	MNGR_INT	R/W	х	х	х	х	х	Х	х	х			
one.	TOWN		х	CLR_ FAST	CLR_R	RINT[1:0]	CLR_PEDGE	CLR_SAMP SAMP_IT[1:0]					
			FAST[1:0]		FAST_TH[5:0]							
0x05	MNGR_ DYN	R/W			BLOFF_HI_IT[7:0]								
	J					BLOFF_LO_IT[7:0]							
0x08	SW_RST	W			D	Data Required for Execution = 0x000000							
0x09	SYNCH	W			Data Required for Execution = 0x000000								
0x0A	FIFO_RST	W			Data Required for Execution = 0x000000								
			0	0 1 0 1			REV_	ID[3:0]					
0x0F	INFO	R	Х	х	0	1	х	Х	Х	Х			
			Х	х	х	Х	х	Х	Х	х			
			EN_ULP_L	.ON[1:0]	FMSTR[1:0]		EN_ECG	EN_BIOZ	EN_PACE	х			
0x10	CNFG_GEN	R/W	EN_BLOF	F[1:0]	EN_DC	LOFF[1:0]	IPOL		IMAG[2:0]				
			VTH[1	:0]	EN_RI	BIAS[1:0]	RBIAS	SV[1:0]	RBIASP	RBIASN			
	CNEC		Х	EN_VCAL	VMODE	VMAG	х	Х	Х	Х			
0x12	CNFG_ CAL	R/W	Х		FCAL[2:0]		FIFTY		THIGH[10:8]				
	-					Т	HIGH[7:0]						
0.44	CNFG_	DAM	ECG_POL	х	ECG_ OPENP	ECG_ OPENN	ECG_CAL	P_SEL[1:0]	ECG_CALI	N_SEL[1:0]			
0x14	EMUX	R/W	Х	х	x	х	х	Х	х	х			
			Х	х	х	х	х	х	х	х			
			ECG_RA	ΓΕ[1:0]	х	х	х	Х	ECG_G	AIN[1:0]			
0x15	CNFG_ ECG	R/W	х	ECG_ DHPF	ECG_[DLPF[1:0]	Х	Х	х	x			
			Х	х	х	х	х	х	х	х			

User Command and Register Map (continued)

REG		R/W				D/	ATA INDEX			
[6:0]	NAME	MODE	23/15/7	22/14/6	21/13/5	20/12/4	19/11/3	18/10/2	17/9/1	16/8/0
			х	х	BMUX_ OPENP	BMUX_ OPENN	BMUX_CALP_SEL[1:0] BMUX_CALN_SEL[N_SEL[1:0]
0x17	CNFG_ BMUX	R/W	х	х	BMUX_CG_MODE[1:0]		BMUX_EN_ BIST	BMUX_RNOM[2:0])]
			Х	ВІ	MUX_RMOD	[2:0]	х	Х	BMUX_FI	BIST[1:0]
			BIOZ_RATE	Е	BIOZ_AHPF[2:0]	EXT_RBIAS	LN_BIOZ	BIOZ_G	AIN[1:0]
0x18	CNFG_	R/W	BIOZ_DHI	PF[1:0]	BIOZ_I	DLPF[1:0]		BIOZ_F0	CGEN[3:0]	
0.10	BioZ	1000	BIOZ_ CGMON	ВІ	OZ_CGMAG	6[2:0]		BIOZ_PI	HOFF[3:0]	
			PACE_POL	х	х	х	DIFF_OFF		PACE_GAIN[2:0]	
0x1A	CNFG_ PACE	R/W	х	AOUT_ AOUT[1:0] _		x	x	x	x	
				PACE_DA	ACP[3:0]			PACE_D	DACN[3:0]	
				WNDV	V[3:0]			RGA	IN[3:0]	
0x1D	CNFG_ RTOR1	R/W	EN_RTOR	х	PAV	/G[1:0]	PTSF[3:0]			
	KIOKI		Х	х	х	х	х	Х	Х	х
			Х	х			Н	OFF[5:0]	,	
0x1E	0x1E		Х	х	RAV	/G[1:0]	х		RHSF[2:0]	
	TOTAL		Х	х	х	х	х	Х	х	х
0x20	ECG_FIFO_ BURST	R+	ECG	FIFO Burst N	Mode Read E	Back	See FIFO Description for details			
0x21	ECG_FIFO	R	ECG F	IFO Normal	Mode Read	Back	See FIFO Description for details			
0x22	BIOZ_ FIFO_ BURST	R+	BioZ	FIFO Burst N	Mode Read B	Back	See FIFO Description for details			
0x23	BIOZ_FIFO	R	BioZ F	IFO Normal	Mode Read	Back	See FIFO Description for details			
0x25	RTOR	R	R-to-F	R Interval Re	gister Read I	Back		See FIFO Desc	ription for details	
0x30	PACE0_ BURST	R	PACE0 (Data	Sets 0 to 5)	Burst Mode	Read Back		See PACE Desc	cription for details	3
0x31	PACE0_A	R	PACE0 (Data S	Sets 0 and 1)	Normal Mod	de Read Back		See PACE Desc	cription for details	;
0x32	PACE0_B	R	PACE0 (Data 9	Sets 2 and 3)	Normal Mod	de Read Back	See PACE Description for details			
0x33	PACE0_C	R	PACE0 (Data S	Sets 4 and 5)	Normal Mod	de Read Back		See PACE Desc	cription for details	3
0x34	PACE1_ BURST	R	PACE1 (Data	Sets 0 to 5)	Burst Mode	Read Back	See PACE Description for details			
0x35	PACE1_A	R	PACE1 (Data S	Sets 0 and 1)	Normal Mod	de Read Back	See PACE Description for details			
0x36	PACE1_B	R	PACE1 (Data S	Sets 2 and 3)	Normal Mod	de Read Back	See PACE Description for details			
0x37	PACE1_C	R	PACE1 (Data S	Sets 4 and 5)	Normal Mod	de Read Back		See PACE Desc	cription for details	;

User Command and Register Map (continued)

REG	NAME	R/W				DA	TA INDEX			
[6:0]	NAME	MODE	23/15/7	22/14/6	21/13/5	20/12/4	19/11/3	18/10/2	17/9/1	16/8/0
0x38	PACE2_ BURST	R+	PACE2 (Data	Sets 0 to 5)	Burst Mode	Read Back	See PACE Description for details			
0x39	PACE2_A	R	PACE2 (Data S	Sets 0 and 1)	Normal Mod	le Read Back	See PACE Description for details			
0x3A	PACE2_B	R	PACE2 (Data S	Sets 2 and 3)	Normal Mod	le Read Back		See PACE Desc	cription for details	
0x3B	PACE2_C	R	PACE2 (Data S	Sets 4 and 5)	Normal Mod	le Read Back		See PACE Desc	cription for details	
0x3C	PACE3_ BURST	R+	PACE3 (Data	Sets 0 to 5)	Burst Mode	Read Back		See PACE Desc	cription for details	
0x3D	PACE3_A	R	PACE3 (Data S	Sets 0 and 1)	Normal Mod	le Read Back		See PACE Desc	cription for details	
0x3E	PACE3_B	R	PACE3 (Data S	Sets 2 and 3)	Normal Mod	le Read Back		See PACE Desc	cription for details	
0x3F	PACE3_C	R	PACE3 (Data S	Sets 4 and 5)	Normal Mod	le Read Back		See PACE Desc	cription for details	
0x40	PACE4_ BURST	R+	PACE4 (Data	Sets 0 to 5)	Burst Mode	Read Back		See PACE Desc	cription for details	
0x41	PACE4_A	R	PACE4 (Data S	Sets 0 and 1)	Normal Mod	le Read Back		See PACE Desc	cription for details	
0x42	PACE4_B	R	PACE4 (Data S	Sets 2 and 3)	Normal Mod	le Read Back		See PACE Desc	cription for details	
0x43	PACE4_C	R	PACE4 (Data S	Sets 4 and 5)	Normal Mod	le Read Back		See PACE Desc	cription for details	
0x44	PACE5_ BURST	R+	PACE5 (Data S	Sets 0 to 5) B	Burst Mode R	ead Back		See PACE Desc	cription for details	
0x45	PACE5_A	R	PACE5 (Data S	Sets 0 and 1)	Normal Mod	le Read Back		See PACE Desc	cription for details	
0x46	PACE5_B	R	PACE5 (Data S	Sets 2 and 3)	Normal Mod	le Read Back		See PACE Desc	cription for details	
0x47	PACE5_C	R	PACE5 (Data S	PACE5 (Data Sets 4 and 5) Normal Mode Read Back See PACE Description for details						
0x7F	NO-OP	R/W	x/x/x	x/x/x	x/x/x	x/x/x	x/x/x	x/x/x	x/x/x	x/x/x

Note: R/W Mode R+ denotes burst mode.

x = Don't Care

Register Description

NO_OP (0x00 and 0x7F) Registers

No Operation (NO_OP) registers are read-write registers that have no internal effect on the device. If these registers are read back, DOUT remains zero for the entire SPI transaction. Any attempt to write to these registers is ignored without impact to internal operation.

STATUS (0x01) Register

STATUS is a read-only register that provides a comprehensive overview of the current status of the device. The first two bytes indicate the state of all interrupt bits (regardless of whether interrupts are enabled in registers EN_INT (0x02) or EN_INT2 (0x03)). All interrupt bits are active high. The last byte includes detailed status information for conditions associated with the other interrupt bits.

Table 11. STATUS (0x01) Register Map

REG	NAME	R/W	23/15/7	22/14/6	21/13/5	20/12/4	19/11/3	18/10/2	17/9/1	16/8/0
			EINT	EOVF	FSTINT	DCLOFFINT	BINT	BOVF	BOVER	BUNDR
0x01	01 STATUS R	BCGMON	PINT	POVF	PEDGE	LONINT	RRINT	SAMP	PLLINT	
0,01		х	х	BCGMP	BCGMN	LDOFF_ PH	LDOFF_ PL	LDOFF_ NH	LDOFF_ NL	

Table 12. Status (0x01) Register Meaning

INDEX	NAME	MEANING
D[23]	EINT	ECG FIFO Interrupt. Indicates that ECG records meeting/exceeding the ECG FIFO Interrupt Threshold (EFIT) are available for readback. Remains active until ECG FIFO is read back to the extent required to clear the EFIT condition.
D[22]	EOVF	ECG FIFO Overflow. Indicates that the ECG FIFO has overflown and the data record has been corrupted. Remains active until a FIFO Reset (recommended) or SYNCH operation is issued.
D[21]	FSTINT	ECG Fast Recovery Mode. Issued when the ECG Fast Recovery Mode is engaged (either manually or automatically). Status and Interrupt Clear behavior is defined by CLR_FAST, see MNGR_INT for details.
D[20]	DCLOFFINT	DC Lead-Off Detection Interrupt. Indicates that the MAX30001 has determined it is in an ECG leads off condition (as selected in CNFG_GEN) for more than 90ms. Remains active as long as the leads-off condition persists, then held until cleared by STATUS read back (32nd SCLK).
D[19]	BINT	BioZ FIFO Interrupt. Indicates BioZ records meeting/exceeding the BioZ FIFO Interrupt Threshold (BFIT) are available for read back. Remains active until BioZ FIFO is read back to the extent required to clear the BFIT condition.
D[18]	BOVF	BioZ FIFO Overflow. Indicates the BioZ FIFO has overflowed and the data record has been corrupted. Remains active until a FIFO Reset (recommended) or SYNCH operation is issued.
D[17]	BOVER	BioZ Over Range. Indicates the BioZ output magnitude has exceeded the BioZ High Threshold (BLOFF_HI_IT) for at least 100ms, recommended for use in 2 and 4 electrode BioZ Lead Off detection. Remains active as long as the condition persists, then held until cleared by STATUS read back (32nd SCLK).
D[16]	BUNDR	BioZ Under Range. Indicates the BioZ output magnitude has been bounded by the BioZ Low Threshold (BLOFF_LO_IT) for at least 1.7 seconds, recommended for use in 4 electrode BioZ Lead Off detection. Remains active as long as the condition persists, then held until cleared by STATUS read back (32nd SCLK).

Table 12. Status (0x01) Register Meaning (continued)

INDEX	NAME	MEANING
D[15]	BCGMON	BioZ Current Generator Monitor. Indicates the DRVP and/or DRVN current generator has been in a Lead Off condition for at least 128ms, recommended for use in 4 electrode BioZ Lead Off detection. Remains active as long as the condition persists, then held until cleared by STATUS read back (32nd SCLK).
D[14]	PINT	PACE FIFO Interrupt. Indicates PACE records are available for read back (should be used in conjunction with EINT). Remains active until all available PACE FIFO records have been read back.
D[13]	POVF	PACE FIFO Overflow. Indicates the PACE FIFO has overflowed and the data record has been corrupted. Remains active until a FIFO Reset (recommended) or SYNCH operation is issued.
D[12]	PEDGE	PACE Edge Detection Interrupt. Real time PACE edge indicator showing when the MAX30001 has determined a PACE edge occurred (note this is different than the PINT interrupt, which indicates when the detected edges are logged into the PACE FIFO). Clear behavior is defined by CLR_PEDGE[1:0], see the MNGR_INT (0x04) register for details.
D[11]	LONINT	Ultra-Low Power (ULP) Leads-On Detection Interrupt. Indicates that the MAX30001 has determined it is in a leads-on condition (as selected in CNFG_GEN). LONINT is asserted whenever EN_ULP_LON[1:0] in register CNFG_GEN is set to either 01 or 10 to indicate that the ULP leads on detection mode has been enabled. The STATUS register has to be read back once after ULP leads on detection mode has been activated to clear LONINT and enable leads on detection. LONINT remains active while the leads-on condition persists, then held until cleared by STATUS read back (32nd SCLK).
D[10]	RRINT	ECG R-to-R Detector R Event Interrupt. Issued when the R-to-R detector has identified a new R event. Clear behavior is defined by CLR_RRINT[1:0]; see MNGR_INT for details.
D[9]	SAMP	Sample Synchronization Pulse. Issued on the ECG base-rate sampling instant, for use in assisting µC monitoring and synchronizing other peripheral operations and data, generally recommended for use as a dedicated interrupt. Frequency is selected by SAMP_IT[1:0], see MNGR_INT for details. Clear behavior is defined by CLR_SAMP, see MNGR_INT for details.
D[8]	PLLINT	PLL Unlocked Interrupt. Indicates that the PLL has not yet achieved or has lost its phase lock. PLLINT will only be asserted when the PLL is powered up and active (ECG and/or BioZ Channel enabled). Remains asserted while the PLL unlocked condition persists, then held until cleared by STATUS read back (32nd SCLK).
D[5]	BCGMP	BioZ Current Generator Monitor Positive Output. Indicates the DRVP current generator has been in a Lead Off condition for at least 128ms. This is not strictly an interrupt bit, but is a detailed status bit, covered by the BCGMON interrupt bit.
D[4]	BCGMN	BioZ Current Generator Monitor Negative Output. Indicates the DRVN current generator has been in a Lead Off condition for at least 128ms. This is not strictly an interrupt bit, but is a detailed status bit, covered by the BCGMON interrupt bit.
D[3]	LDOFF_PH	DC Lead Off Detection Detailed Status. Indicates that the MAX30001 has determined (as selected by
D[2]	LDOFF_PL	CNFG_GEN): ECGP is above the high threshold (V _{THH}), ECGP is below the low threshold (V _{THL}), ECGN is above
D[1]	LDOFF_NH	the high threshold (VT _{HH}), ECGN is below the low threshold (V _{THL}), respectively. Remains active as long as the leads-off detection is active and the leads-off condition persists, then
D[0]	LDOFF_NL	held until cleared by STATUS read back (32nd SCLK). LDOFF_PH to LDOFF_NL are detailed status bits that are asserted at the same time as DCLOFFINT.

EN_INT (0x02) and EN_INT2 (0x03) Registers

EN_INT and EN_INT2 are read/write registers that govern the operation of the INTB output and INT2B output, respectively. The first two bytes indicate which interrupt input bits are included in the interrupt output OR term (ex. a one in an EN_INT register indicates that the corresponding input bit is included in the INTB interrupt output OR term). See the STATUS register for detailed descriptions of the interrupt bits. The power-on reset state of all EN_INT bits is 0 (ignored by INT).

EN_INT and EN_INT2 can also be used to mask persistent interrupt conditions in order to perform other interrupt-driven operations until the persistent conditions are resolved.

INTB_TYPE[1:0] allows the user to select between a CMOS or an open-drain NMOS mode INTB output. If using open-drain mode, an option for an internal $125k\Omega$ pullup resistor is also offered.

All INTB and INT2B types are active-low (INTB low indicates the device requires servicing by the μ C); however, the open-drain mode allows the INTB line to be shared with other devices in a wired-or configuration.

In general, it is suggested that INT2B be used to support specialized/dedicated interrupts of use in specific applications, such as the self-clearing versions of SAMP or RRINT.

Table 13. EN INT (0x02) and EN INT2 (0x03) Register Maps

REG	NAME	R/W	23/15/7	22/14/6	21/13/5	20/12/4	19/11/3	18/10/2	17/9/1	16/8/0
			EN_EINT	EN_ EOVF	EN_ FSTINT	EN_DCL OFFINT	EN_BINT	EN_BOVF	EN_ BOVER	EN_ BUNDR
0x02 0x03	EN_INT EN_INT2	R/W	EN_ BCGMON	EN_PINT	EN_POVF	EN_ PEDGE	EN_ LONINT	EN_ RRINT	EN_ SAMP	EN_ PLLINT
			x	Х	х	Х	Х	х	INTB_T	YPE[1:0]

Table 14. EN_INT (0x02 and 0x03) Register Meaning

INDEX	NAME	DEFAULT	FUNCTION
D[23:8]	EN_EINT EN_ EOVF EN_FSTINT EN_DCLOFFINT EN_BINT EN_BOVF EN_BOVER EN_BUNDR EN_BCGMON EN_PINT EN_POVF EN_PEDGE EN_LONINT EN_ RRINT EN_SAMP EN_PLLINT	0x0000	Interrupt Enables for interrupt bits in STATUS[23:8] 0 = Individual interrupt bit is not included in the interrupt OR term 1 = Individual interrupt bit is included in the interrupt OR term
		11	INTB Port Type (EN_INT Selections) 00 = Disabled (high impedance) 01 = CMOS Driver 10 = Open-Drain NMOS Driver 11 = Open-Drain NMOS Driver with Internal 125kΩ Pullup Resistance
D[1:0]	INTB_TYPE[1:0]	11	INT2B Port Type (EN_INT2 Selections) 00 = Disabled (high impedance) 01 = CMOS Driver 10 = Open-Drain nMOS Driver 11 = Open-Drain nMOS Driver with Internal 125kΩ Pullup Resistance

MNGR_INT (0x04)

MNGR_INT is a read/write register that manages the operation of the configurable interrupt bits in response to ECG and BioZ FIFO conditions (see the STATUS register and ECG and BioZ FIFO descriptions for more details). Finally, this register contains the configuration bits supporting the sample synchronization pulse (SAMP) and RTOR heart rate detection interrupt (RRINT).

Table 15. MNGR_INT (0x04) Register Map

REG	NAME	R/W	23/15/7	22/14/6	21/13/5	20/12/4	19/11/3	18/10/2	17/9/1	16/8/0
					EFIT[4:0]				BFIT[2:0]	
0x04	MNGR_	R- R/W	х	Х	х	х	х	х	х	Х
0,04	INT		х	CLR_ FAST	CLR_RF	RINT[1:0]	CLR_ PEDGE	CLR_ SAMP	SAMP_	_IT[1:0]

Table 16. MNGR_INT (0x04) Register Functionality

INDEX	NAME	DEFAULT	FUNCTION
D[23:19]	EFIT[4:0]	01111	ECG FIFO Interrupt Threshold (issues EINT based on number of unread FIFO records) 00000 to 11111 = 1 to 32, respectively (i.e. EFIT[4:0]+1 unread records)
D[18:16]	BFIT[2:0]	011	BioZ FIFO Interrupt Threshold (issues BINT based on number of unread FIFO records) 000 to 111 = 1 to 8, respectively (i.e. BFIT[2:0]+1 unread records)
D[6]	CLR_FAST	0	FAST MODE Interrupt Clear Behavior: 0 = FSTINT remains active until the FAST mode is disengaged (manually or automatically), then held until cleared by STATUS read back (32nd SCLK). 1 = FSTINT remains active until cleared by STATUS read back (32nd SCLK), even if the MAX30001 remains in FAST recovery mode. Once cleared, FSTINT will not be re-asserted until FAST mode is exited and re-entered, either manually or automatically.
D[5:4]	CLR_RRINT[1:0]	00	RTOR R Detect Interrupt (RRINT) Clear Behavior: 00 = Clear RRINT on STATUS Register Read Back 01 = Clear RRINT on RTOR Register Read Back 10 = Self-Clear RRINT after one ECG data rate cycle, approximately 2ms to 8ms 11 = Reserved. Do not use.
D[3]	CLR_PEDGE	0	PACE Edge Detect Interrupt (PEDGE) Clear Behavior 0 = Clear PEDGE on STATUS Register Read Back 1 = Self-Clear PEDGE after one PACE comparison cycle, roughly 16µs Note: Self-Clear mode is recommended for INT2B use only.
D[2]	CLR_SAMP	1	Sample Synchronization Pulse (SAMP) Clear Behavior: 0 = Clear SAMP on STATUS Register Read Back (recommended for debug/ evaluation only). 1 = Self-clear SAMP after approximately one-fourth of one data rate cycle.
D[1:0]	SAMP_IT[1:0]	00	Sample Synchronization Pulse (SAMP) Frequency 00 = issued every sample instant 01 = issued every 2nd sample instant 10 = issued every 4th sample instant 11 = issued every 16th sample instant

MNGR_DYN (0x05)

MNGR_DYN is a read/write register that manages the settings of any general/dynamic modes within the device. The ECG Fast Recovery modes and thresholds are managed here. This register also contains the interrupt thresholds for BioZ AC Lead-Off Detection (see CNFG_GEN for more details). Unlike many CNFG registers, changes to dynamic modes do not impact FIFO operations or require a SYNCH operation (though the affected circuits may require time to settle, resulting in invalid/corrupted FIFO output voltage information during the settling interval).

Table 17. MNGR_DYN (0x05) Register Map

REG	NAME	R/W	23/15/7	22/14/6	21/13/5	20/12/4	19/11/3	18/10/2	17/9/1	16/8/0	
			FAST[1:0]		FAST_TH[5:0]						
0x05	MNGR_ DYN	R/W		BLOFF_HI_IT[7:0]							
	DIN					BLOFF_L	O_IT[7:0]				

Table 18. MNGR_DYN (0x05) Register Functionality

INDEX	NAME	DEFAULT	FUNCTION
D[23:22]	FAST[1:0]	00	ECG Channel Fast Recovery Mode Selection (ECG High Pass Filter Bypass): 00 = Normal Mode (Fast Recovery Mode Disabled) 01 = Manual Fast Recovery Mode Enable (remains active until disabled) 10 = Automatic Fast Recovery Mode Enable (Fast Recovery automatically activated when/while ECG outputs are saturated, using FAST_TH). 11 = Reserved. Do not use.
D[21:16]	FAST_TH[5:0]	0x3F	Automatic Fast Recovery Threshold: If FAST[1:0] = 10 and the output of an ECG measurement exceeds the symmetric thresholds defined by 2048*FAST_TH for more than 125ms, the Fast Recovery mode will be automatically engaged and remain active for 500ms. For example, the default value (FAST_TH = 0x3F) corresponds to an ECG output upper threshold of 0x1F800, and an ECG output lower threshold of 0x20800.
D[15:8]	BLOFF_HI_IT[7:0]	0xFF	BioZ AC Lead Off Over-Range Threshold If EN_BLOFF[1:0] = 1x and the ADC output of a BioZ measurement exceeds the symmetric thresholds defined by ±2048*BLOFF_HI_IT for over 128ms, the BOVER interrupt bit will be asserted. For example, the default value (BLOFF_IT= 0xFF) corresponds to a BioZ output upper threshold of 0x7F800 or about 99.6% of the full scale range, and a BioZ output lower threshold of 0x80800 or about 0.4% of the full scale range with the LSB weight ≈ 0.4%.
D[7:0]	BLOFF_LO_IT[7:0]	0xFF	BioZ AC Lead Off Under-Range Threshold If EN_BLOFF[1:0] = 1x and the output of a BioZ measurement is bounded by the symmetric thresholds defined by ±32*BLOFF_LO_IT for over 128ms, the BUNDR interrupt bit will be asserted.

SW_RST (0x08)

SW_RST (Software Reset) is a write-only register/command that resets the MAX30001 to its original default conditions at the end of the SPI SW_RST transaction (i.e. the 32nd SCLK rising edge). Execution occurs only if DIN[23:0] = 0x0000000. The effect of a SW_RST is identical to power-cycling the device.

Table 19. SW RST (0x08) Register Map

REG	NAME	R/W	23/15/7	22/14/6	21/13/5	20/12/4	19/11/3	18/10/2	17/9/1	16/8/0
						D[23:16] = 0x00			
0x08	SW_RST	W				D[15:8]	= 0x00			
						D[7:0]	= 0x00			

SYNCH (0x09)

SYNCH (Synchronize) is a write-only register/command that begins new ECG/BioZ operations and recording, beginning on the internal MSTR clock edge following the end of the SPI SYNCH transaction (i.e. the 32nd SCLK rising edge). Execution occurs only if DIN[23:0] = 0x0000000. In addition to resetting and synchronizing the operations of any active ECG, RtoR, BioZ, and PACE circuitry, SYNCH will also reset and clear the FIFO memories and the DSP filters (to midscale), allowing the user to effectively set the "Time Zero" for the FIFO records. No configuration settings are impacted. For best results, users should wait until the PLL has achieved lock before synchronizing if the CNFG_GEN settings have been altered.

Once the device is initially powered up, it will need to be fully configured prior to launching recording operations. Likewise, anytime a change to CNFG_GEN, CNFG_ ECG, or CNFG_BIOZ registers are made there may be discontinuities in the ECG and BioZ records and possibly changes to the size of the time steps recorded in the FIFOs. The SYNCH command provides a means to restart operations cleanly following any such disturbances.

During multi-channel operations, if a FIFO overflow event occurs and a portion of the record is lost, it is recommended to use the SYNCH command to recover and restart the recording (avoiding issues with missing data in one or more channel records). Note that the two channel records cannot be directly synchronized within the device, due to significant differences in group delays, depending on filter selections—alignment of the records will have to be done externally.

Table 20. SYNCH (0x09) Register Map

REG	NAME	R/W	23/15/7	22/14/6	21/13/5	20/12/4	19/11/3	18/10/2	17/9/1	16/8/0
						D[23:16] = 0x00			
0x09	SYNCH	W				D[15:8]	= 0x00			
						D[7:0]	= 0x00			

FIFO_RST (0x0A)

FIFO_RST (FIFO Reset) is a write-only register/command that begins a new ECG and BioZ recordings by resetting the FIFO memories and resuming the record with the next available ECG and BioZ data. Execution occurs only if DIN[23:0] = 0x000000. Unlike the SYNCH command, the operations of any active ECG, R-to-R, BioZ, and PACE circuitry are not impacted by FIFO_RST, so no settling/recovery transients apply. FIFO_RST can also be used to quickly recover from a FIFO overflow state (recommended for single ECG or BioZ channel use, see above).

Table 21. FIFO RST (0x0A) Register Map

REG	NAME	R/W	23/15/7	22/14/6	21/13/5	20/12/4	19/11/3	18/10/2	17/9/1	16/8/0
						D[23:16]] = 0x00			
0x0A	FIFO_RST	W				D[15:8]	= 0x00			
						D[7:0]	= 0x00			

INFO (0x0F)

INFO is a read-only register that provides information about the MAX30001. The first nibble contains an alternating bit pattern to aide in interface verification. The second nibble contains the revision ID. The third nibble includes part ID information.

Note: Due to internal initialization procedures, this command will not read-back valid data if it is the first command executed following either a power-cycle event, or a SW_RST event.

Table 22. INFO (0x0F) Register Map

REG	NAME	R/W	23/15/7	22/14/6	21/13/5	20/12/4	19/11/3	18/10/2	17/9/1	16/8/0
		O R	0	1	0	1	REV_ID[3:0]			
0x0F	INFO		Х	Х	0	1	х	Х	х	х
			х	х	х	х	х	х	х	х

Table 23. INFO (0x0F) Register Meaning

INDEX	NAME	MEANING
D[19:16]	REV_ID[3:0]	Revision ID

CNFG_GEN (0x10)

CNFG_GEN is a read/write register which governs general settings, most significantly the master clock rate for all internal timing operations. Anytime a change to CNFG_GEN is made, there may be discontinuities in the ECG and BioZ records and possibly changes to the size of the time steps recorded in the FIFOs. The SYNCH command can be used to restore internal synchronization resulting from configuration changes. Note when EN_ECG and EN_BIOZ are both logic-low, the device is in one of two ultra-low power modes (determined by EN_ULP_LON).

Table 24. CNFG_GEN (0x10) Register Map

REG	NAME	R/W	23/15/7	22/14/6	21/13/5	20/12/4	19/11/3	18/10/2	17/9/1	16/8/0
	ONEO		EN_ULP_	EN_ULP_LON[1:0]		FMSTR[1:0]		EN_BIOZ EN_PACE		Х
0x10	0x10 CNFG_ GEN R/W	R/W	EN_BLOFF[1:0]		EN_DCLOFF[1:0]		IPOL	IMAG[2:0]		
	OLIV		VTH	[1:0]	EN_RBIAS[1:0]		RBIAS	SV[1:0]	RBIASP	RBIASN

Table 25. CNFG_GEN (0x10) Register Functionality

INDEX	NAME	DEFAULT	FUNCTION
D[23:22]	EN_ULP_LON [1:0]	00	Ultra-Low Power Lead-On Detection Enable 00 = ULP Lead-On Detection disabled 01 = ECG ULP Lead-On Detection enabled 10 = Reserved. Do not use. 11 = Reserved. Do not use. ULP mode is only active when the ECG channel is powered down/disabled.

Table 25. CNFG_GEN (0x10) Register Functionality (continued)

INDEX	NAME	DEFAULT	FUNCTION
D[21:20]	FMSTR[1:0]	00	Master Clock Frequency. Selects the Master Clock Frequency (FMSTR), and Timing Resolution (T_{RES}), which also determines the ECG and CAL timing characteristics. These are generated from FCLK, which is always 32.768kHz. 00 = F_{MSTR} = 32768Hz, T_{RES} = 15.26 μ s (512Hz ECG progressions) 01 = F_{MSTR} = 32000Hz, T_{RES} = 15.63 μ s (500Hz ECG progressions) 10 = F_{MSTR} = 32000Hz, T_{RES} = 15.63 μ s (200Hz ECG progressions) 11 = F_{MSTR} = 31968.78Hz, T_{RES} = 15.64 μ s (199.8049Hz ECG progressions)
D[19]	EN_ECG	0	ECG Channel Enable 0 = ECG Channel disabled 1 = ECG Channel enabled Note: The ECG channel must be enabled to allow R-to-R operation.
D[18]	EN_BIOZ	0	BioZ Channel Enable 0 = BioZ Channel disabled 1 = BioZ Channel enabled
D[17]	EN_PACE	0	PACE Channel Enable 0 = PACE Channel disabled 1 = PACE Channel enabled if ECG channel also enabled (EN_ECG=1)
D[15:14]	EN_BLOFF[1:0]	00	BioZ Digital Lead Off Detection Enable 00 = Digital Lead Off Detection disabled 01 = Lead Off Under Range Detection, 4 electrode BioZ applications 10 = Lead Off Over Range Detection, 2 and 4 electrode BioZ applications 11 = Lead Off Over & Under Range Detection, 4 electrode BioZ applications AC Method, requires active BioZ Channel , enables BOVER & BUNDR interrupt behavior. Uses BioZ excitation current set in CNFG_BIOZ with digital thresholds set in MNGR_DYN.
D[13:12]	EN_DCLOFF	00	DC Lead-Off Detection Enable 00 = DC Lead-Off Detection disabled 01 = DCLOFF Detection applied to the ECGP/N pins 10 = Reserved. Do not use. 11 = Reserved. Do not use. DC Method, requires active selected channel, enables DCLOFF interrupt and status bit behavior. Uses current sources and comparator thresholds set below.
D[11]	DCLOFF_IPOL	0	DC Lead-Off Current Polarity (if current sources are enabled/connected) 0 = ECGP - Pullup
D[10:8]	IMAG[2:0]	000	DC Lead-Off Current Magnitude Selection 000 = 0nA (Disable and Disconnect Current Sources) 001 = 5nA 010 = 10nA 011 = 20nA 100 = 50nA 101 = 100nA 111 = Reserved. Do not use. 111 = Reserved. Do not use.

Table 25. CNFG_GEN (0x10) Register Functionality (continued)

INDEX	NAME	DEFAULT	FUNCTION
D[7:6]	VTH[1:0]	00	DC Lead-Off Voltage Threshold Selection $00 = V_{MID} \pm 300 \text{mV}$ $01 = V_{MID} \pm 400 \text{mV}$ $10 = V_{MID} \pm 450 \text{mV}$ $11 = V_{MID} \pm 500 \text{mV}$
D[5:4]	EN_RBIAS[1:0]	00	Enable and Select Resistive Lead Bias Mode 00 = Resistive Bias disabled 01 = ECG Resistive Bias enabled if EN_ECG is also enabled 10 = BioZ Resistive Bias enabled if EN_BIOZ is also enabled 11 = Reserved. Do not use. If EN_ECG or EN_BIOZ is not asserted at the same time or prior to EN_RBIAS[1:0] being enabled, then EN_RBIAS[1:0] will remain set to 00.
D[3:2]	RBIASV[1:0]	01	Resistive Bias Mode Value Selection $00 = R_{BIAS} = 50 M\Omega$ $01 = R_{BIAS} = 100 M\Omega$ $10 = R_{BIAS} = 200 M\Omega$ $11 = Reserved. Do not use.$
D[1]	RBIASP	0	Enables Resistive Bias on Positive Input $0 = ECGP/BIP$ is not resistively connected to V_{MID} $1 = ECGP/BIN$ is connected to V_{MID} through a resistor (selected by RBIASV).
D[0]	RBIASN	0	Enables Resistive Bias on Negative Input $0 = ECGN$ is not resistively connected to V_{MID} $1 = ECGN$ is connected to V_{MID} through a resistor (selected by RBIASV).

Table 26 shows the ECG and BioZ data rates that can be realized with various setting of FMSTR, along with RATE configuration bits available in the CNFG_ECG and CNFG_BIOZ registers. Note FMSTR also determines the timing resolution of the PACE detection block (and the resulting record depth with respect to the ECG_RATE selection) as well as the timing resolution of the CAL waveform generator.

Table 26. Master Frequency Summary Table

FMSTR [1:0]	MASTER FREQUENCY (f _{MSTR}) (Hz)	ECG DATA RATE (ECG_RATE) (sps)	RTOR TIMING RESOLUTION (RTOR_RES) (ms)	PACE TIMING RESOLUTION (PACE_RES) (µs)	PACE FIFO RECORD DEPTH (ECG_RATE)	CALIBRATION TIMING RESOLUTION (CAL_RES) (μs)	BioZ DATA RATES (B_RATE) (sps)
00	32,768	00 = 512 01 = 256 10 = 128	7.8125	15.26	00 = 128 01 = 256 1x = 512	30.52	0 = 64 1 = 32
01	32,000	00 = 500 01 = 250 10 = 125	8.000	15.63	00 = 128 01 = 256 1x = 512	31.25	0 = 62.50 1 = 31.25
10	32,000	10 = 200	8.000	15.63	320	31.25	0 = 50 1 = 25
11	31,968	10 = 199.8049	8.008	15.64	320	31.28	0 = 49.95 1 = 24.98

CNFG_CAL (0x12)

CNFG_CAL is a read/write register that configures the operation, settings, and function of the Internal Calibration Voltage Sources (VCALP and VCALN). The output of the voltage sources can be routed to the ECG or BioZ/PACE inputs through the channel input MUXes to facilitate end-to-end testing operations. Note if a VCAL source is applied to a connected device, it is recommended that the appropriate channel MUX switches be placed in the OPEN position.

Table 27. CNFG_CAL (0x12) Register Map

REG	NAME	R/W	23/15/7	22/14/6	21/13/5	20/12/4	19/11/3	18/10/2	17/9/1	16/8/0	
	01150		Х	EN_VCAL	VMODE	VMAG	х	Х	х	х	
0x12	CNFG_ CAL	R/W	Х		FCAL[2:0]		FIFTY	THIGH[10:8]			
	OAL		THIGH[7:0]								

Table 28. CNFG_CAL (0x12) Register Functionality

INDEX	NAME	DEFAULT	FUNCTION
D[22]	EN_VCAL	0	Calibration Source (VCALP and VCALN) Enable 0 = Calibration sources and modes disabled 1 = Calibration sources and modes enabled
D[21]	VMODE	0	Calibration Source Mode Selection 0 = Unipolar, sources swing between V _{MID} ± V _{MAG} and V _{MID} 1 = Bipolar, sources swing between V _{MID} + V _{MAG} and V _{MID} - V _{MAG}
D[20]	VMAG	0	Calibration Source Magnitude Selection (V _{MAG}) 0 = 0.25mV 1 = 0.50mV
D[14:12]	FCAL[2:0]	100	Calibration Source Frequency Selection (FCAL) $000 = F_{MSTR}/128 \qquad (256, 250, \text{ or } 249.75\text{Hz}) \\ 001 = F_{MSTR}/512 \qquad (64, 62.5, \text{ or } 62.4375\text{Hz}) \\ 010 = F_{MSTR}/2048 \qquad (16, 15.625, \text{ or } 15.609375\text{Hz}) \\ 011 = F_{MSTR}/8192 \qquad (4, 3.90625, \text{ or } 3.902344\text{Hz}) \\ 100 = F_{MSTR}/2^{15} \qquad (1, 0.976563, \text{ or } 0.975586\text{Hz}) \\ 101 = F_{MSTR}/2^{17} \qquad (0.25, 0.24414, \text{ or } 0.243896\text{Hz}) \\ 110 = F_{MSTR}/2^{19} \qquad (0.0625, 0.061035\text{Hz}, \text{ or } 0.060974\text{Hz}) \\ 111 = F_{MSTR}/2^{21} \qquad (0.015625, 0.015259, \text{ or } 0.015244\text{Hz}) \\ \text{Actual frequencies are determined by FMSTR selection (see CNFG_GEN for details), frequencies in parenthesis are based on 32,768, 32,000, or 31,968\text{Hz} clocks (FMSTR[1:0] = 00). TCAL = 1/FCAL. }$
D[11]	FIFTY	1	Calibration Source Duty Cycle Mode Selection 0 = Use CAL_THIGH to select time high for VCALP and VCALN 1 = THIGH = 50% (CAL_THIGH[10:0] are ignored)
D[10:0]	THIGH[10:0]	0x000	Calibration Source Time High Selection If FIFTY = 1, t _{HIGH} = 50% (and THIGH[10:0] are ignored), otherwise THIGH = THIGH[10:0] x CAL_RES CAL_RES is determined by FMSTR selection (see CNFG_GEN for details); for example, if FMSTR[1:0] = 00, CAL_RES = 30.52µs.

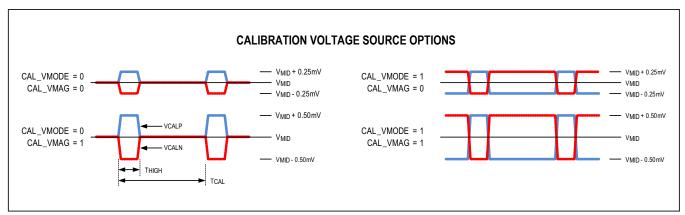


Figure 16. Calibration Voltage Source Options

CNFG_EMUX (0x14)

CNFG_EMUX is a read/write register which configures the operation, settings, and functionality of the Input Multiplexer associated with the ECG channel.

Table 29. CNFG_EMUX (0x14) Register Map

REG	NAME	R/W	23/15/7	22/14/6	21/13/5	20/12/4	19/11/3	18/10/2	17/9/1	16/8/0
0.44	CNFG	G	ECG_POL	х	ECG_ OPENP	ECG_ OPENN	ECG_CAL	P_SEL[1:0]	ECG_CALN	_SEL[1:0]
0x14	0x14 EMUX F	R/W	Х	х	х	х	х	х	Х	х
			Х	х	х	х	х	х	Х	х

Table 30. CNFG_EMUX (0x14) Register Functionality

INDEX	NAME	DEFAULT	FUNCTION
D[23]	ECG_POL	0	ECG Input Polarity Selection 0 = Non-inverted 1 = Inverted
D[21]	ECG_OPENP	1	Open the ECGP Input Switch (most often used for testing and calibration) 0 = ECGP is internally connected to the ECG AFE Channel 1 = ECGP is internally isolated from the ECG AFE Channel
D[20]	ECG_OPENN	1	Open the ECGN Input Switch (most often used for testing and calibration) 0 = ECGN is internally connected to the ECG AFE Channel 1 = ECGN is internally isolated from the ECG AFE Channel
D[19:18]	ECG_CALP_ SEL[1:0]	00	ECGP Calibration Selection 00 = No calibration signal applied 01 = Input is connected to V _{MID} 10 = Input is connected to VCALP (only available if CAL_EN_VCAL = 1) 11 = Input is connected to VCALN (only available if CAL_EN_VCAL = 1)
D[17:16]	ECG_CALN_ SEL[1:0]	00	ECGN Calibration Selection 00 = No calibration signal applied 01 = Input is connected to V _{MID} 10 = Input is connected to VCALP (only available if CAL_EN_VCAL = 1) 11 = Input is connected to VCALN (only available if CAL_EN_VCAL = 1)

CNFG_ECG (0x15)

CNFG_ECG is a read/write register which configures the operation, settings, and functionality of the ECG channel. Anytime a change to CNFG_ECG is made, there may be discontinuities in the ECG record and possibly changes to the size of the time steps recorded in the ECG FIFO. The SYNCH command can be used to restore internal synchronization resulting from configuration changes.

Table 31. CNFG_ECG (0x15) Register Map

REG	NAME	R/W	23/15/7	22/14/6	21/13/5	20/12/4	19/11/3	18/10/2	17/9/1	16/8/0
		CNFG_ R/W	ECG_F	RATE[1:0]	х	х	Х	Х	ECG_GA	AIN[1:0]
0x15			Х	ECG_DHPF	ECG_DI	LPF[1:0]	Х	Х	Х	х
			Х	х	х	х	Х	Х	Х	х

Table 32. CNFG_ECG (0x15) Register Functionality

INDEX	NAME	DEFAULT	FUNCTION
			ECG Data Rate (also dependent on FMSTR selection, see CNFG_GEN Table 33):
			FMSTR = 00: f _{MSTR} = 32768Hz, t _{RES} = 15.26μs (512Hz ECG progressions) 00 = 512sps 01 = 256sps 10 = 128sps 11 = Reserved. Do not use. FMSTR = 01: f _{MSTR} = 32000Hz, t _{RES} = 15.63μs (500Hz ECG progressions) 00 = 500sps 01 = 250sps 10 = 125sps 11 = Reserved. Do not use.
D[23:22]	ECG_RATE[1:0]	10	FMSTR = 10: f _{MSTR} = 32000Hz, t _{RES} = 15.63µs (200Hz ECG progressions) 00 = Reserved. Do not use. 01 = Reserved. Do not use. 10 = 200sps 11 = Reserved. Do not use. FMSTR = 11: f _{MSTR} = 31968Hz, t _{RES} = 15.64µs (199.8Hz ECG progressions)
			00 = Reserved. Do not use. 01 = Reserved. Do not use. 10 = 199.8sps 11 = Reserved. Do not use.

Table 32. CNFG_ECG (0x15) Register Functionality (continued)

INDEX	NAME	DEFAULT	FUNCTION
D[17:16]	ECG_GAIN[1:0]	00	ECG Channel Gain Setting 00 = 20V/V 01 = 40V/V 10 = 80V/V 11 = 160V/V
D[14]	ECG_DHPF	1	ECG Channel Digital High-Pass Filter Cutoff Frequency 0 = Bypass (DC) 1 = 0.50Hz
D[13:12]	ECG_DLPF[1:0]	01	ECG Channel Digital Low-Pass Filter Cutoff Frequency 00 = Bypass (Decimation only, no FIR filter applied) 01 = approximately 40Hz (Except for 125 and 128sps settings, see Table 33) 10 = approximately 100Hz (Available for 512, 256, 500, and 250sps ECG Rate selections only) 11 = approximately 150Hz (Available for 512 and 500sps ECG Rate selections only) Note: See Table 33. If an unsupported DLPF setting is specified, the 40Hz setting (ECG_DLPF[1:0] = 01) will be used internally; the CNFG_ECG register will continue to hold the value as written, but return the effective internal value when read back.

Table 33. Supported ECG_RATE and ECG_DLPF Options

CNFG_GEN	ECG_RATE[1:0]	ECG_DLPF[1:0]/DIGITAL LPF CUTOFF					
FMSTR[1:0]	SAMPLE RATE (sps)	00	01 (Hz)	10 (Hz)	11 (Hz)		
	00 = 512	Bypass	40.96	102.4	153.6		
00 = 32,768Hz	01 = 256	Bypass	40.96	102.4	40.96		
	10 = 128	Bypass	28.35	28.35	28.35		
	00 = 500	Bypass	40.00	100.0	150.0		
01 = 32,000Hz	01 = 250	Bypass	40.00	100.0	40.00		
	10 = 125	Bypass	27.68	27.68	27.68		
10 = 32,000Hz	10 = 200	Bypass	40.00	40.00	40.00		
11 = 31,968Hz	10 = 199.8	Bypass	39.96	39.96	39.96		

Note: Combinations shown in grey are unsupported and will be internally mapped to the default settings shown.

CNFG_BMUX(0x17)

CNFG_BMUX is a read/write register which configures the operation, settings, and functionality of the input multiplexer associated with the BioZ channel.

Table 34. CNFG_BMUX (0x17) Register Map

REG	NAME	R/W	23/15/7	22/14/6	21/13/5	20/12/4	19/11/3	18/10/2	17/9/1	16/8/0
	2015		х	х	BMUX_OPENP	BMUX_ OPENN	BMUX_CAL	P_SEL[1:0]	BMUX_CAL	N_SEL[1:0]
0x17	BMUX x x		х	х	BMUX_CG_MC	DDE[1:0]	BMUX_ EN_BIST	ВМ	MUX_RNOM[2	:0]
			BMUX_RMOD[2:0]		Х	Х	BMUX_F	BIST[1:0]		

Table 35. CNFG_BMUX (0x17) Register Functionality

INDEX	NAME	DEFAULT	FUNCTION
D[21]	BMUX_ OPENP	1	Open the BIP Input Switch (most often used for testing and calibration) 0 = BIP is internally connected to the BioZ channel 1 = BIP is internally isolated from the BioZ channel
D[20]	BMUX_ OPENN	1	Open the BIN Input Switch (most often used for testing and calibration) 0 = BIN is internally connected to the BioZ channel 1 = BIN is internally isolated from the BioZ channel
D[19:18]	BMUX_CALP_ SEL[1:0]	00	BIP Calibration Selection (VCAL application to BIP/N inputs intended for use in PACE testing only.) 00 = No calibration signal applied 01 = Input is connected to VMID 10 = Input is connected to VCALP (only available if CAL_EN_VCAL=1) 11 = Input is connected to VCALN (only available if CAL_EN_VCAL=1)
D[17:16]	BMUX_CALN_ SEL[1:0]	00	BIN Calibration Selection (VCAL application to BIP/N inputs intended for use in PACE testing only.) 00 = No calibration signal applied 01 = Input is connected to VMID 10 = Input is connected to VCALP (only available if CAL_EN_VCAL=1) 11 = Input is connected to VCALN (only available if CAL_EN_VCAL=1)
D[13:12]	BMUX_CG_ MODE[1:0]	00	BioZ Current Generator Mode Selection 00 = Unchopped Sources with Low Pass Filter (higher noise, excellent 50/60Hz rejection, recommended for ECG, BioZ applications) 01 = Chopped Sources without Low Pass Filter (low noise, no 50/60Hz rejection, recommended for BioZ applications with digital LPF, possibly battery powered ECG, BioZ applications) 10 = Chopped Sources with Low Pass Filter (low noise, excellent 50/60Hz rejection) 11 = Chopped Sources with Resistive CM Setting (Not recommended to be used for drive currents >32μA) (low noise, excellent 50/60Hz rejection, lower input impedance)

Table 35. CNFG_BMUX (0x17) Register Functionality (continued)

INDEX	NAME	DEFAULT	FUNCTION
D[11]	BMUX_EN_ BIST	0	BioZ Modulated Resistance Built-In-Self-Test (RMOD BIST) Mode Enable 0 = RMOD BIST Disabled 1 = RMOD BIST Enabled Note: Available only when CNFG_CAL EN_VCAL= 0 To avoid body interference, the BIP/N switches should be open in this mode. When enabled, the DRVP/N isolation switches are opened and the DRVP/N-to-BIP/N internal switches are engaged. Also, the lead bias resistors are applied to the BioZ inputs in $200\text{M}\Omega$ mode.
D[10:8]	BMUX_ RNOM[2:0]	000	BioZ RMOD BIST Nominal Resistance Selection See RMOD BIST Settings Table for details.
D[6:4]	BMUX_ RMOD[2:0]	100	BioZ RMOD BIST Modulated Resistance Selection (See RMOD BIST Settings Table for details.) 000 = Modulated Resistance Value 0 001 = Modulated Resistance Value 1 010 = Modulated Resistance Value 2 011 = Reserved, Do Not Use 1xx = All SWMOD Switches Open - No Modulation (DC value = RNOM)
D[1:0]	BMUX_ FBIST[1:0]	00	BioZ RMOD BIST Frequency Selection Calibration Source Frequency Selection (FCAL) 00 = f _{MSTR} /2 ¹³ (Approximately 4 Hz) 01 = f _{MSTR} /2 ¹⁵ (Approximately 1 Hz) 10 = f _{MSTR} /2 ¹⁷ (Approximately 1/4 Hz) 11 = f _{MSTR} /2 ¹⁹ (Approximately 1/16 Hz) Actual frequencies are determined by FMSTR selection (see CNFG_GEN for details), approximate frequencies are based on a 32,768 Hz clock (FMSTR[1:0]=00). All selections use 50% duty cycle.

Table 36. CNFG_BMUX (0x17) RMOD BIST Settings

BMUX_RNOM[2:0]	BMUX_RMOD[2:0]	NOMINAL RESISTANCE (Ω)	MODULATED RESISTANCE (mΩ)
000	000 001 010 1xx	5000	2960.7 980.6 247.5 Unmodulated
001	000 001 010 1xx	2500	740.4 245.2 61.9 Unmodulated
010	000 001 010 1xx	1667	329.1 109.0 27.5 Unmodulated

Table 36. CNFG_BMUX (0x17) RMOD BIST Settings (continued)

BMUX_RNOM[2:0] AND SWNOM SWITCHES ENGAGED	BMUX_RMOD[2:0]	NOMINAL RESISTANCE (Ω)	MODULATED RESISTANCE (mΩ)
011	000 001 1xx	1250	185.1 61.3 Unmodulated
100	000 001 1xx	1000	118.5 39.2 Unmodulated
101	000 001 1xx	833	82.3 27.2 Unmodulated
110	000 001 1xx	714	60.5 20.0 Unmodulated
111	000 001 1xx	625	46.3 15.3 Unmodulated

CNFG_BIOZ(0x18)

CNFG_BIOZ is a read/write register which configures the operation, settings, and function of the BioZ channel, including the associated modulated current generator. Anytime a change to CNFG_BIOZ is made, there may be discontinuities in the BioZ record and possibly changes to the size of the time steps recorded in the BioZ FIFO. The SYNCH command can be used to restore internal synchronization resulting from configuration changes.

Table 37. CNFG_BIOZ (0x18) Register Map

REG	NAME	R/W	23/15/7	22/14/6	21/13/5	20/12/4	19/11/3	18/10/2	17/9/1	16/8/0
		CNFG_ R/W	BIOZ_ RATE	BIOZ_AHPF[2:0]			EXT_ RBIAS	LN_BIOZ	BIOZ_G	AIN[1:0]
0x18	CNFG_ BioZ		BIOZ_DF	IPF[1:0]	BIOZ_DLI	PF[1:0]		BIOZ_FC	GEN[3:0]	
	5102		BIOZ_ CGMON	ı	BIOZ_CGMAG[2	:0]		BIOZ_PF	IOFF[3:0]	

Table 38. CNFG_BIOZ (0x18) Register Functionality

INDEX	NAME	DEFAULT	FUNCTION
			BioZ Data Rate (also dependent on FMSTR selection, see CNFG_GEN):
			FMSTR = 00: f _{MSTR} = 32,768Hz (512Hz ECG/BioZ progressions) 0 = 64sps 1 = 32sps
D[23]	BIOZ_RATE	0	FMSTR = 01: f _{MSTR} = 32,000Hz (500Hz ECG/BioZ progressions) 0 = 62.50sps 1 = 31.25sps
			FMSTR = 10: f _{MSTR} = 32,000 Hz (200Hz ECG/BioZ progressions) 0 = 50sps 1 = 25sps
			FMSTR = 11: f _{MSTR} = 31,968 Hz (199.8Hz ECG/BioZ progressions) 0 = 49.95sps 1 = 24.98sps
D[22:20]	BIOZ_ AHPF[2:0]	010	BioZ/PACE Channel Analog High-Pass Filter Cutoff Frequency and Bypass 000 = 125Hz 001 = 300Hz 010 = 800Hz 011 = 2000Hz 100 = 3700Hz 101 = 7200Hz 11x = Bypass AHPF
D[19]	EXT_RBIAS	0	External Resistor Bias Enable 0 = Internal Bias Generator used 1 = External Bias Generator used Note: Use of the external resistor bias will improve the temperature coefficient of all biases within the product, but the main benefit is improved control of BioZ current generator magnitude. If enabled, the user must include the required external resistor between R _{BIAS} and GND, and the temperature coefficent achieved will be determined by the combined performance of the internal bandgap and the external resistor.
D[18]	LN_BIOZ	0	BioZ Channel Instrumentation Amplifier (INA) Power Mode 0 = BioZ INA is in low power mode 1 = BioZ INA is in low noise mode
D[17:16]	BIOZ_ GAIN[1:0]	00	BioZ Channel Gain Setting 00 = 10V/V 01 = 20V/V 10 = 40V/V 11 = 80V/V
D[15:14]	BIOZ_ DHPF[1:0]	00	BioZ Channel Digital High-Pass Filter Cutoff Frequency 00 = Bypass (DC) 01 = 0.05Hz 1x = 0.50Hz

Table 38. CNFG_BIOZ (0x18) Register Functionality (continued)

INDEX	NAME	DEFAULT	FUNCTION
D[13:12]	BIOZ_ DLPF[1:0]	01	BioZ Channel Digital Low-Pass Filter Cutoff Frequency 00 = Bypass (Decimation only, no FIR filter) 01 = 4Hz 10 = 8Hz 11 = 16Hz (Available for 64, 62.5, 50, and 49.95sps BioZ Rate selections only) Note: See Table 39 below. If an unsupported DLPF setting is specified, the 4Hz setting (BIOZ_DLPF[1:0] = 01) will be used internally; the CNFG_BIOZ register will continue to hold the value as written, but return the effective internal value when read back.
D[11:8]	BIOZ_ FCGEN[3:0]	1000	BioZ Current Generator Modulation Frequency $0000 = 4*f_{MSTR}$ (approximately 128000Hz) $1000 = f_{MSTR}/64$ (approximately 500Hz) $0001 \approx 2*f_{MSTR}$ (approximately 80000Hz) $1001 = f_{MSTR}/128$ (approximately 250Hz) $0010 \approx f_{MSTR}$ (approximately 40000Hz) $101x = f_{MSTR}/256$ (approximately 125Hz) $0011 \approx f_{MSTR}/2$ (approximately 18000Hz) $11xx = f_{MSTR}/256$ (approximately 125Hz) $0100 = f_{MSTR}/4$ (approximately 8000Hz) $0101 = f_{MSTR}/8$ (approximately 4000Hz) $0110 = f_{MSTR}/8$ (approximately 2000Hz) $0111 = f_{MSTR}/32$ (approximately 1000Hz) Actual frequencies determined by FMSTR selection, see CNFG_GEN register and table below for details. Frequencies expected between approximately 16kHz and approximately 64kHz are offset to approximately 18kHz to approximately 80kHz to reduce ECG/PACE channel crosstalk. PACE operation is only supported at approximately 40kHz and approximately 80kHz offset selections: FCGEN[3:0] = 0001,0010, at other selections, PACE will be rendered inoperable.
D[7]	BIOZ_ CGMON	0	BioZ Current Generator Monitor 0 = Current Generator Monitors disabled 1 = Current Generator Monitors enabled, requires active BioZ channel and Current Generators. Enables BCGMON interrupt and status bit behavior. Monitors current source compliance levels, useful in detecting DRVP/DRVN lead off conditions with 4 electrode BioZ applications.
D[6:4]	BIOZ_ CGMAG[2:0]	000	BioZ Current Generator Magnitude 000 = Off (DRVP and DRVN floating, Current Generators Off) 001 = 8μA 010 = 16μA 011 = 32μA 100 = 48μA 101 = 64μA 111 = 96μA See Table 40 and 41 below for a list of allowed BIOZ_CGMAG settings vs. FCGEN selections.
D[3:0]	BIOZ_ PHOFF[3:0]	0000	BioZ Current Generator Modulation Phase Offset Phase Resolution and Offset depends on BIOZ_FCGEN setting: BIOZ_FCGEN[3:0] ≥ 0010: Phase Offset = BIOZ_PHOFF[3:0]*11.25° (0 to 168.75°) BIOZ_FCGEN[3:0] = 0001: Phase Offset = BIOZ_PHOFF[3:1]*22.50° (0 to 157.50°) BIOZ_FCGEN[3:0] = 0000: Phase Offset = BIOZ_PHOFF[3:2]*45.00° (0 to 135.00°)

Table 39. Supported BIOZ_RATE and BIOZ_DLPF Options

CNFG_GEN	BIOZ_RATE		BIOZ_DLPF[1:0] / Digital LPF Cut Off									
FMSTR[1:0]	Sample Rate	00	01	10	11							
00 = 22.76947	0 = 64sps	Pyroco	4.096Hz	8.192Hz	16.384Hz							
00 = 32,768Hz	1 = 32sps	Bypass	4.09002	0.19202	4.096Hz							
04 - 22 00011-	0 = 62.5sps	Pyroco	4.0Hz	8.0Hz	16.0Hz							
01 = 32,000Hz	1 = 31.25sps	Bypass	4.002	0.0П2	4.0Hz							
10 = 32.000Hz	0 = 50sps	Pyroco	4.0Hz	8.0Hz	16.0Hz							
10 = 32,000H2	1 = 25sps	Bypass	4.0HZ	8.UHZ	4.0Hz							
44 04 00011	0 = 49.95sps	Dymana	3.996Hz	7 00011-	15.984Hz							
11 = 31,968Hz	1 = 25.98sps	Bypass	J.990FZ	7.992Hz	3.996Hz							

Note: Combinations shown in grey are unsupported and will be internally mapped to the default settings shown.

Table 40. Actual BioZ Current Generator Modulator Frequencies vs. FMSTR[1:0] Selection

		BioZ Current Generator Modulation Frequency (Hz)											
BIOZ_FCGEN[3:0]	FMSTR[1:0] = 00 f _{MSTR} = 32,768Hz	FMSTR[1:0] = 01 f _{MSTR} = 32,000Hz	FMSTR[1:0] = 10 f _{MSTR} = 32,000Hz	FMSTR[1:0] = 11 f _{MSTR} = 31,968Hz									
0000	131,072	128,000	128,000	127,872									
0001	81,920	80,000	80,000	81,920									
0010	40,960	40,000	40,000	40,960									
0011	18,204	17,780	17,780	18,204									
0100	8,192	8,000	8,000	7,992									
0101	4,096	4,000	4,000	3,996									
0110	2,048	2,000	2,000	1,998									
0111	1,024	1,000	1,000	999									
1000	512	500	500	500									
1001	256	250	250	250									
101x, 11xx	128	125	125	125									

Note: Shaded selections are intentionally offset to improve ECG/PACE system crosstalk.

Table 41. Allowed CGMAG Option vs. FCGEN Selections

FCGEN[3:0]	APPROXIMATE CURRENT GENERATOR MODULATION FREQUENCY (Hz)	CGMAG[2:0] OPTIONS ALLOWED	CURRENT GENERATOR MAGNITUDE OPTIONS ALLOWED (μΑ _{P-P})					
0000	12,8000							
0001	80,000	All	All					
0010	40,000	All	All					
0011	18,000							
0100	8,000	All except 111	All except 96					
0101	4,000	000, 001, 010, 011	Off, 8, 16, 32					
0110	2,000	000, 001, 010	Off, 8, 16					
0111	1,000							
1000	500	000 004	0# 0					
1001	250	000, 001	Off, 8					
101x, 11xx	125							

CNFG_PACE (0x1A) Register

CNFG_PACE is a read/write register which configures the operation, settings, and function of the PACE detection channel. Portions of the PACE AFE are shared with the BioZ channel so anytime a change to CNFG_BIOZ or CNFG_PACE is made, there may be discontinuities in the combined ECG/PACE FIFO output. The SYNCH command can be used to restore internal synchronization resulting from configuration changes.

Note if enabling the PACE function, the Analog High-Pass Filter in the shared BioZ/PACE AFE must be set to the desired value via BIOZ_AHPF[1:0] in the CNFG_BIOZ register, even if the BioZ function is disabled (EN_BIOZ = 0 in CNFG_GEN register.

Table 42. CNFG_PACE (0x1A) Register Map

REG	NAME	R/W	23/15/7	22/14/6	21/13/5	20/12/4	19/11/3	18/10/2	17/9/1	16/8/0		
0.44	0x1A CNFG_ PACE	D 444	PACE_ POL	х	х	х	DIFF_OFF	PACE_GAIN[2:0]				
UX1A		R/W	х	AOUT_LBW	AOU ⁻	T[1:0]	х	Х	Х	х		
				PACE_DACP[3	:0]	PACE_DACN[3:0]						

Table 43. CNFG_PACE (0x1A) Register Functionality

INDEX	NAME	DEFAULT		FUNCTION							
D[23]	PACE_POL	0	PACE Input Polarity Selection 0 = Non-Inverted 1 = Inverted								
D[19]	DIFF_OFF	0	PACE Differentiator (Derivative) 0 = Enable Differentiator functio 1 = Disable Differentiator function	n (default)	ction						
D[18:16]	PACE_ GAIN[2:0]	000	PACE Channel Gain Selection Normal Mode (AOUT = 00) 000 = 45*4*3 = 540 001 = 45*2*3 = 270 010 = 20*4*3 = 240 011 = 20*2*3 = 120 100 = 5*4*3 = 60 101 = 5*2*3 = 30 110 = 2.2*4*3 = 26.4 111 = 2.2*2*3 = 13.2	INA OUT Mode (AOUT = 01) 45*1.125 = 50.625 45*1.125 = 22.500 20*1.125 = 22.500 5*1.125 = 5.625 5*1.125 = 5.625 5*1.125 = 5.625 5*1.125 = 5.625	PGA OUT Mode (AOUT = 10) 45*4*1.125 = 202.50 45*2*1.125 = 101.25 20*4*1.125 = 90.00 20*2*1.125 = 45.00 5*4*1.125 = 22.50 5*2*1.125 = 11.25 5*4*1.125 = 22.50 5*2*1.125 = 11.25						
D[14]	AOUT_LBW	0	PACE Analog Output Buffer Bar 0 = Maximum BW (approximate 1 = Limited BW (approximately This selection is only relevant w	ely 100kHz) 16kHz)	AOUT≠00.						
D[13:12]	AOUT[1:0]	00		10 = PACE PGA Output							
D[7:4]	PACE_ DACP[3:0]	0101	PACE Detector Positive Compa VDACP = PACE_DACP[3:0]*22								
D[3:0]	PACE_ DACN[3:0]	0101	PACE Detector Negative Comp VDACN = -PACE_DACN[3:0]*2								

CNFG_RTOR1 and CNFG_RTOR2 (0x1D and 0x1E)

CNFG_RTOR is a two-part read/write register that configures the operation, settings, and function of the R-to-R heart rate detection block. The first register contains algorithmic voltage gain and threshold parameters, the second contains algorithmic timing parameters.

Table 44. CNFG_RTOR1 and CNFG_RTOR2 (0x1D and 0x1E) Register Maps

REG	NAME	R/W	23/15/7	22/14/6	21/13/5	20/12/4	19/11/3	18/10/2	17/9/1	16/8/0			
		R/W		WNDW[3	3:0]		RGAIN[3:0]						
0x1D	0x1D CNFG_ RTOR1		EN_ RTOR	х	PAVO	G[1:0]	PTSF[3:0]						
			Х	х	х	х	х	x x x					
	01150		Х	х									
0x1E	0x1E CNFG_ RTOR2		Х	Х	RAV	G[1:0]	х	x RHSF[2:0]					
			Х	x	х	х	х	х	Х	х			

Table 45. CNFG_RTOR1 (0x1D) Register Functionality

INDEX	NAME	DEFAULT	FUNCTION
CNFG_RT	OR1 (0x1D)		
D[23:20]	WNDW[3:0]	0011	This is the width of the averaging window, which adjusts the algorithm sensitivity to the width of the QRS complex. R-to-R Window Averaging (Window Width = WNDW[3:0]*8ms) 0000 = 6 x RTOR_RES 0001 = 8 x RTOR_RES 0010 = 10 x RTOR_RES 0011 = 12 x RTOR_RES 0011 = 12 x RTOR_RES 0100 = 14 x RTOR_RES 0110 = 16 x RTOR_RES 0110 = 18 x RTOR_RES 0111 = 20 x RTOR_RES 1000 = 22 x RTOR_RES 1001 = 24 x RTOR_RES 1011 = 28 x RTOR_RES 1011 = 28 x RTOR_RES 1100 = Reserved. Do not use. 1110 = Reserved. Do not use. 1111 = Reserved. Do not use. The value of RTOR_RES is approximately 8ms, see Table 26.
D[19:16]	RGAIN[3:0]	1111	R-to-R Gain (where Gain = 2^RGAIN[3:0], plus an auto-scale option). This is used to maximize the dynamic range of the algorithm. 0000 = 1

Table 45. CNFG_RTOR1 (0x1D) Register Functionality (continued)

INDEX	NAME	DEFAULT	FUNCTION
D[15]	EN_RTOR	0	ECG R-to-R Detection Enable 0 = R-to-R Detection disabled 1 = R-to-R Detection enabled if EN_ECG is also enabled.
D[13:12]	PAVG[1:0]	10	R-to-R Peak Averaging Weight Factor This is the weighting factor for the current R-to-R peak observation vs. past peak observations when determining peak thresholds. Lower numbers weight current peaks more heavily. 00 = 2 01 = 4 10 = 8 (default) 11 = 16 Peak_Average(n) = [Peak(n) + (PAVG-1) x Peak_Average(n-1)] / PAVG.
D[11:8]	PTSF[3:0]	0011	R-to-R Peak Threshold Scaling Factor This is the fraction of the Peak Average value used in the Threshold computation. Values of 1/16 to 16/16 are selected by (PTSF[3:0]+1)/16, default is 4/16.

Table 46. CNFG_RTOR2 (0x1E) Register Functionality

CNFG_RT	OR2 (0x1E)		
D [21:16]	HOFF[5:0] 10_0000		R-to-R Minimum Hold Off This sets the absolute minimum interval used for the static portion of the Hold Off criteria. Values of 0 to 63 are supported, default is 32 $t_{HOLD_OFF_MIN} = HOFF[5:0] * t_{RTOR}, \text{ where } t_{RTOR} \text{ is approximately 8ms, as determined by FMSTR[1:0] in the CNFG_GEN register.} $ (representing approximately ½ second). The R-to-R Hold Off qualification interval is $t_{Hold_Off} = MAX(t_{Hold_Off_Min}, t_{Hold_Off_Dyn}) \text{ (see below)}.$
D[13:12]	RAVG[1:0]	10	R-to-R Interval Averaging Weight Factor This is the weighting factor for the current R-to-R interval observation vs. the past interval observations when determining dynamic holdoff criteria. Lower numbers weight current intervals more heavily. 00 = 2 01 = 4 10 = 8 (default) 11 = 16 Interval_Average(n) = [Interval(n) + (RAVG-1) x Interval_Average(n-1)] / RAVG.
D[10:8]	RHSF[2:0]	100	R-to-R Interval Hold Off Scaling Factor This is the fraction of the R-to-R average interval used for the dynamic portion of the holdoff criteria (t _{HOLD_OFFDYN}). Values of 0/8 to 7/8 are selected by RTOR_RHSF[3:0]/8, default is 4/8. If 000 (0/8) is selected, then no dynamic factor is used and the holdoff criteria is determined by HOFF[5:0] only (see above).

FIFO Memory Description

The device provides read only FIFO memory for ECG, BioZ, and PACE information. A single memory register is also supported for heart rate detection output data (R-to-R). The operation of these FIFO memories and registers is detailed in the following sections.

<u>Table 47</u> summarizes the method of access and data structure within the FIFO memory.

ECG FIFO Memory (32 Words x 24 Bits)

The ECG FIFO memory is a standard circular FIFO consisting of 32 words, each with 24 bits of information.

The ECG FIFO is independently managed by internal read and write pointers. The read pointer is updated in response to the 32nd SCLK rising edge in a normal mode read back transaction and on the (32 + n x 24)th SCLK rising edge(s) in a burst mode transaction where n = 0 to up to 31. Once a FIFO sample is marked as read, it cannot be accessed again.

The write pointer is governed internally. To aide data management and reduce μC overhead, the device provides a user-programmable ECG FIFO Interrupt Threshold (EFIT[4:0]) governing the ECG interrupt bit (EINT). This threshold can be programmed with values from 1 to 32, representing the number of unread ECG FIFO entries required before the EINT bit will be asserted, alerting the μC that there is a significant amount of data in the ECG FIFO ready for read back (see MNGR_INT (0x04) for details).

Do not read beyond the last valid FIFO word to prevent possible data corruption.

If the write pointer ever traverses the entire FIFO array and catches up to the read pointer (due to failure of the μ C to read/maintain FIFO data), a FIFO overflow will occur and data will be corrupted. The EOVF STATUS and tag bits will indicate this condition and the FIFO should be cleared before continuing measurements using either a SYNCH or FIFO_RST command—note overflow events will result in the loss of samples and thus timing information, so these conditions should not occur in well-designed applications.

Table 47. FIFO Memory Access and Data Structure Summary

DEC	FIFO		DATA STRUCTURE (D[23:0])																						
REG	AND MODE	23	22	21	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
0x20	ECG Burst		ECG Sample Voltage Data [17:0] ETA [2:													ETA([2:0			TAC 2:0]	-					
0x21	ECG		ECG Sample Voltage Data [17:0] ETA [2:												ETA([2:0			TAC 2:0]	-						
		23	22	21	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
0x22	BioZ Burst							BioZ	. Sam	nple V	oltage	Data	19:0)]								0		TAC 2:0]	-
0x23	BioZ							BioZ	. Sam	nple V	oltage	Data	ı [19:0	0]								0		TAC 2:0]	-
		23	22	21	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
0x25	RTOR					RTOF	R Inte	val Ti	ming	Data	[13:0]					0	0	0	0	0	0	0	0	0	0

ECG FIFO Data Structure

The data portion of the word contains the 18-bit ECG voltage information measured at the requested sample rate in left justified two's complement format. The remaining six bits of data hold important data tagging information (see details in Table 48 and Table 49).

After converting the data portion of the sample to signed magnitude format, the ECG input voltage is calculated by the following equation:

 V_{ECG} (mV) = ADC x V_{REF} / (2¹⁷ x ECG_GAIN) where:

ADC = ADC counts in signed magnitude format, V_{REF} = 1000mV (typ) (refer to the *Electrical Characteristics* section), and ECG_GAIN = 20V/V, 40V/V, 80V/V, or 160V/V, set in CNFG_ECG (0x15).

ECG Data Tags (ETAG)

Three bits in the sample record are used as an ECG data tag (ETAG[2:0] = D[5:3]). This section outlines the meaning of the various data tags used in the ECG FIFO

and recommended handling within the continuous ECG record.

VALID: ETAG = 000 indicates that ECG data for this sample represents both a valid voltage and time step in the ECG record.

FAST: ETAG = 001 indicates that ECG data for this sample was taken in the FAST settling mode and that the voltage information in the sample should be treated as transient and invalid. Note that while the voltage data is invalid, samples of this type do represent valid time steps in the ECG record.

VALID EOF: ETAG = 010 indicates that ECG data for this sample represents both a valid voltage and time step in the ECG record, and that this is the last sample currently available in the ECG FIFO (End-of-File, EOF). The μ C should wait until further samples are available before requesting more data from the ECG FIFO.

FAST EOF: ETAG = 011 indicates that ECG data for this sample was taken in the FAST settling mode and that the voltage information in the sample should be treated as transient and invalid. Note that while the voltage data is

Table 48. ECG FIFO - ECG Data Tags (ETAG[2:0] = D[5:3])

ETAG [2:0]	MEANING	DETAILED DESCRIPTION	RECOMMENDED USER ACTION	DATA VALID	TIME VALID
000	Valid Sample	This is a valid FIFO sample.	Log sample into ECG record and increment the time step. Continue to gather data from the ECG FIFO.	Yes	Yes
001	Fast Mode Sample	This sample was taken while the ECG channel was in a FAST recovery mode. The voltage information is not valid, but the sample represents a valid time step.	Discard, note, or post-process this voltage sample, but increment the time base. Continue to gather data from the ECG FIFO.	No	Yes
010	Last Valid Sample (EOF)	This is a valid FIFO sample, but this is the last sample currently available in the FIFO (End of File indicator).	Log sample into ECG record and increment the time step. Suspend read back operations on the ECG FIFO until more samples are available.	Yes	Yes
011	Last Fast Mode Sample (EOF)	See above (ETAG=001), but in addition, this is the last sample currently available in the FIFO (End of File indicator).	Discard, note, or post-process this voltage sample, but increment the time base. Suspend read back operations on the ECG FIFO until more samples are available.	No	Yes
10x	Unused				
110	FIFO Empty (Exception)	This is an invalid sample provided in response to an SPI request to read an empty FIFO.	Discard this sample, without incrementing the time base. Suspend read back operations on this FIFO until more samples are available.	No	No
111	FIFO Overflow (Exception)	The FIFO has been allowed to overflow – the data is corrupted.	Issue a FIFO_RST command to clear the FIFOs or re-SYNCH if necessary. Note the corresponding halt and resumption in ECG/BioZ time/voltage records.	No	No

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invalid, samples of this type do represent valid time steps in the ECG record. In addition, this is the last sample currently available in the ECG FIFO (End-of-File, EOF). The μ C should wait until further samples are available before requesting more data from the ECG FIFO.

EMPTY: ETAG = 110 is appended to any requested read back data from an empty ECG FIFO. The presence of this tag alerts the user that this FIFO data does not represent a valid sample or time step. Note that if handled properly by the μ C, an occurrence of an empty tag will not compromise the integrity of a continuous ECG record — this tag only indicates that the read back request was either premature or unnecessary.

OVERFLOW: ETAG = 111 indicates that the ECG FIFO has overflowed and that there are interruptions or missing data in the sample records. The ECG Overflow (EOVF) bit is also included in the STATUS register. A FIFO_RESET is required to resolve this situation, effectively clearing the FIFO so that valid sampling going forward is assured.

Depending on the application, it may also be necessary to resynchronize the MAX30001 internal channel operations to move forward with valid recordings, the SYNCH command can perform this function while also resetting the FIFO memories.

ECG PACE Data Tag (PTAG)

The PACE FIFO data content is closely linked to ECG FIFO content. If an ECG FIFO samples has related PACE information, this is indicated by a three bit PACE tag (PTAG[2:0] = D[2:0]) appended to and read back at the end of the ECG FIFO sample.

A PACE tag (PTAG) value between 000 and 101 (inclusive) indicates that a PACE event was detected during the sample interval associated with and following the tagged ECG sample. In these cases, PTAG stores a pointer to the appropriate location within the PACE FIFO where the relevant PACE information is stored (see PACE FIFO Memory for more details). A PTAG value of 111 indicates no PACE events were associated with the ECG Sample.

Table 49. ECG FIFO - PACE Data Tags (PTAG[2:0] = D[2:0])

PTAG [2:0]	DETAILED DESCRIPTION	PACE GROUP	RECOMMENDED USER ACTION
000	PACE event detected	0	Associate PACE Group 0 data with this ECG data sample. Follow ETAG recommended user actions.
001	PACE event detected	1	Associate PACE Group 1 data with this ECG data sample. Follow ETAG recommended user actions.
010	PACE event detected	2	Associate PACE Group 2 data with this ECG data sample. Follow ETAG recommended user actions.
011	PACE event detected	3	Associate PACE Group 3 data with this ECG data sample. Follow ETAG recommended user actions.
100	PACE event detected	4	Associate PACE Group 4 data with this ECG data sample. Follow ETAG recommended user actions.
101	PACE event detected	5	Associate PACE Group 5 data with this ECG data sample. Follow ETAG recommended user actions.
110	Unused	-	-
111	No PACE detected	-	Associate PACE Group 0 with this ECG data sample. Follow ETAG recommended user actions.

BioZ FIFO Memory (8 Words x 24 Bits)

The BioZ FIFO memory is a standard circular FIFO consisting of 8 words, each with 24 bits of information. The BioZ FIFO is independently managed by internal read and write pointers. The read pointer is updated in response to the 32nd SCLK rising edge in a normal mode read back transaction and on the $(32 + n \times 24)$ th SCLK rising edge(s) in a burst mode transaction where n = 0 to up to 31. Once a FIFO sample is marked as read, it cannot be accessed again.

The write pointer is governed internally. To aide data management and reduce μC overhead, the device provides a user-programmable BioZ FIFO Interrupt Threshold (BFIT[2:0]) governing the BioZ Interrupt bit (BINT). This threshold can be programmed with values from 1 to 8, representing the number of unread BioZ FIFO entries required before the BINT bit will be asserted, alerting the μC that there is a significant amount of data in the BioZ FIFO ready for read back (see MNGR_INT (0x04) for details).

If the write pointer ever traverses the entire FIFO array and catches up to the read pointer (due to failure of the μ C to read/maintain FIFO data), a FIFO overflow will occur and data will be corrupted. The BOVF STATUS and tag bits will indicate this condition and the FIFO should be cleared before continuing measurements using

either a SYNCH or FIFO_RST command—note overflow events will result in the loss of samples and thus timing information, so these conditions should not occur in well-designed applications.

Do not read beyond the last valid FIFO word to prevent possible data corruption.

BioZ FIFO Data Structure

The data portion of the word contains the 20-bit BioZ voltage information measured at the requested sample rate in left justified two's complement format. One bit is set to 0 and the remaining three bits of data hold important data tagging information (see details in Table 50).

After converting the data portion of the sample to signed magnitude format, BioZ is calculated by the following equation:

BioZ (Ω) = ADC x
$$V_{REF}$$
 / (2¹⁹ x BIOZ_CGMAG x BIOZ_GAIN)

where:

ADC = ADC counts in signed magnitude format, V_{REF} = 1V (typ) (refer to the <u>Electrical Characteristics</u> section), BIOZ_CGMAG = 8 to 96 x 10⁻⁶A, and BIOZ_GAIN = 10V/V, 20V/V, 40V/V, or 80V/V. BIOZ_CGMAG and BIOZ GAIN are set in CNFG_BIOZ (0x18).

Table 50. BioZ FIFO BioZ Data Tags (BTAG[2:0] = D[2:0])

BTAG [2:0]	DESCRIPTION	RECOMMENDED USER ACTION	DATA VALID	TIME VALID
000	Valid Sample	Log sample into BioZ record and increment the time step. Continue to read data from the BioZ FIFO.	Yes	Yes
001	Over/Under Range Sample	Log sample into BioZ record and increment the time step. Determine if the data is valid or a lead off condition. Continue to read data from the BioZ FIFO.	?	Yes
010	Last Valid Sample (EOF)	Log sample into BioZ record and increment the time step. Suspend read of the BioZ FIFO until more samples are available.	Yes	Yes
011	Last Over/Under Range Sample (EOF)	Log sample into BioZ record and increment the time step. Determine if the data is valid or a lead off condition. Suspend read of the BioZ FIFO until more samples are available.	?	Yes
10x	Unused	-	-	-
110	FIFO Empty (exception)	Discard this sample without incrementing the time base. Suspend read of the BioZ FIFO until more samples are available.	No	No
111	FIFO Overflow (exception)	Discard this sample without incrementing the time base. Issue a FIFO_RST command to clear the FIFOs or re-SYNCH if necessary. Note the corresponding halt and resumption in all the FIFOs.	No	No

BioZ Data Tags (BTAG)

The final three bits in the sample are used as a data tag (BTAG[2:0] = D[2:0]) to assist in managing data transfers. The BTAG structure used is detailed below.

VALID: BTAG = 000 indicates that BioZ data for this sample represents both a valid voltage and time step in the BioZ record.

OVER or UNDER RANGE: BTAG = 001 indicates that BioZ data for this sample violated selected range thresholds (see MNGR_DYN and CNFG_GEN) and that the voltage information in the sample should be evaluated to see if it is valid or indicative of a leads-off condition. Note that while the voltage data may be invalid, samples of this type do represent valid time steps in the BioZ record.

VALID EOF: BTAG = 010 indicates that BioZ data for this sample represents both a valid voltage and time step in the BioZ record, and that this is the last sample currently available in the BioZ FIFO (End-of-File, EOF). The μ C should wait until further samples are available before requesting more data from the BioZ FIFO.

OVER or UNDER RANGE EOF: BTAG = 011 indicates that BioZ data for this sample violated selected range thresholds (see MNGR_DYN and CNFG_GEN) and that the voltage information in the sample should be evaluated to see if it is valid or indicates a leads-off condition. Note that while the voltage data may be invalid, samples of this type do represent valid time steps in the BioZ record. This is also the last sample currently available in the BioZ FIFO (End-of-File, EOF). The μ C should wait until further samples are available before requesting more data from the BioZ FIFO.

EMPTY: BTAG = 110 is appended to any requested read back data from an empty BioZ FIFO. The presence of this tag alerts the user that this FIFO data does not represent a valid sample or time step. Note that if handled properly by the μ C, an occurrence of an empty tag will not compromise the integrity of a continuous BioZ record — this tag only indicates that the read back request was either premature or unnecessary.

OVERFLOW: BTAG = 111 indicates that the BioZ FIFO has overflowed and that there are interruptions or missing data in the sample records. The BioZ Overflow (BOVF) bit is also included in the STATUS register. A FIFO_RESET is required to resolve this situation, effectively clearing the FIFO so that valid sampling going forward is assured. Depending on the application, it may also be necessary to resynchronize the MAX30001 internal channel operations to move forward with valid recordings, the SYNCH

command can perform this function while also resetting the FIFO memories.

R-to-R Interval Memory Register (1 Word x 24 Bits)

The R-to-R Interval (RTOR) memory register is a single read-only register consisting of 14 bits of timing interval information, left justified (and 10 unused bits, set to zero).

The RTOR register stores the time interval between the last two R events, as identified by the R-to-R detection circuitry, which operates on the ECG output data. Each LSB in the RTOR register is approximately equal to 8ms (CNFG_GEN for exact figures). The resulting 14-bit storage interval can thus be approximately 130 seconds in length, again depending on device settings.

Each time the R-to-R detector identifies a new R event, the RTOR register is updated, and the RRINT interrupt bit is asserted (see STATUS register for details).

Users wishing to log heart rate based on RTOR register data should set CLR_RRINT equals 01 in the MNGR_INT register. This will clear the RRINT interrupt bit after the RTOR register has been read back, preparing the device for identification of the next R-to-R interval.

Users wishing to log heart rate based on the time elapsed between RRINT assertions using the μ C to keep track of the time base (and ignoring the RTOR register data) have two choices for interrupt management. If CLR_RRINT equals 00 in the MNGR_INT register, the RRINT interrupt bit will clear after each STATUS register read back, preparing the device for identification of the next R-to-R interval. If CLR_RRINT equals 10 in the MNGR_INT register, the RRINT interrupt bit will self-clear after each one full ECG data cycle has passed, preparing the device for identification of the next R-to-R interval (this mode is recommended only if using the INT2B as a dedicated heart rate indicator).

If CLR_RRINT = 0x (interrupt mode) and the R-to-R detector reaches an overflow state after several minutes without detection of an R event, it will assert the RRINT term with a RTOR register value = 0x3FFF, indicating the overflow condition. This interrupt creates a time stamp, allowing the μC to keep track of the time interval between detected R events, even if the signal is lost for a prolonged amount of time. This is important if the RTOR register data is the sole source to keep track of the time base. In the event of an overflow, the RTOR register will be reset after being read back, allowing the μC to track multiple subsequent overflow conditions. RRINT is reset independently of the RTOR register by an appropriate read back operation as specified by the setting of CLR_RRINT.

If CLR_RRINT = 1x (indicator mode) and the R-to-R detector reaches an overflow state after several minutes without detection of an R event, the counter will simply roll over, and the lack of the RRINT activity on the dedicated INT2B line will inform the μC that no R-to-R activity was detected. Generating an interrupt to keep track of the absolute time is not required in this case, as this mode will be used in a system where the μC is used to keep track of the time base.

PACE0 to PACE5 (0x30 to 0x47) Register Groups

The PACE0 to PACE5 register groups are six read only memories used to store pace edge information detected

by the pacemaker detection circuitry. Each pace register group stores data for up to six pace edges detected between two consecutive ECG data samples stored in the ECG_FIFO register and are associated with the leading ECG data sample. The PTAG[2:0] bits for the associated ECG data sample indicate if one or more pace edges were detected and which pace group it was written to. Each pace register group is organized into three subgroup registers denoted by an A, B, or C suffix that are divided into two segments each holding pace edge data for a total of 6 pace edges per group and a grand total of 36 pace edges in 18 registers.

Table 51. PACE0 to PACE5 (0x30 to 0x47) Register Map

REG	NAME	R/W	23 22	21 20) 19	18 1	7 16	15	14	13	12	11	10	9 8	3 7	6	5	4	3 2	2	1	0
0x30	PACE0_BURST	R	,	Bu	rst rea	d of PA	CE0_	A, PA	CE	0_B & PACE	0_C registe	rs (80	0-bit	frame	: 8-	-bit c	omr	nan	d + 3	3*24	4-bit data)	
0x31	PACE0_A	R		PAC	E0_0	DATA[9	:0]	P0_0RFB	P0_0LST	PACE0_1DATA[9:0]						F	P0_1RFB	P0_1LST				
0x32	PACE0_B	R		PAC	E0_2	DATA[9	:0]			P0_2RFB	P0_2LST	PACE0_3DATA[9:0]						F	P0_3RFB	P0_3LST		
0x33	PACE0_C	R		PAC	E0_4	DATA[9	:0]			P0_4RFB	P0_4LST		P	ACE)_5[DATA	A[9:0	0]			P0_5RFB	P0_5LST
0x34	PACE1_BURST	R		Bu	rst rea	d of PA	CE1_	A, PA	CE	1_B & PACE	1_C registe	rs (80	0-bit	frame	e: 8-	-bit c	omr	nan	d + 3	3*24	4-bit data)	
0x35	PACE1_A	R		PAC	E1_0	DATA[9	:0]			P1_0RFB	P1_0LST		P	ACE1	1_1[DATA	۹[9:0	0]		F	P1_1RFB	P1_1LST
0x36	PACE1_B	R		PAC	E1_2	DATA[9	:0]			P1_2RFB	P1_2LST		P	ACE1	I_3I	DATA	A[9:0	0]		F	P1_3RFB	P1_3LST
0x37	PACE1_C	R		PAC	E1_4	DATA[9	:0]			P1_4RFB	P1_4LST		P	ACE1	_5I	DATA	۹[9:0	0]		F	P1_5RFB	P1_5LST
0x38	PACE2_BURST	R		Bu	rst rea	d of PA	CE2_	A, PA	CE	2_B & PACE	2_C registe	rs (80	0-bit	frame	e: 8-	-bit c	omr	nan	d + 3	3*24	4-bit data)	
0x39	PACE2_A	R		PAC	E2_0	DATA[9	:0]			P2_0RFB	P2_0LST		P	ACE2	2_1[DATA	A[9:0	0]		F	P2_1RFB	P2_1LST
0x3A	PACE2_B	R		PAC	E2_2	DATA[9	:0]			P2_2RFB	P2_2LST	PACE2_3DATA[9:0]						F	P2_3RFB	P2_3LST		
0x3B	PACE2_C	R		PAC	E2_4	DATA[9	:0]			P2_4RFB	P2_4LST		P	ACE2	2_51	DATA	A[9:0	0]		F	P2_5RFB	P2_5LST
0x3C	PACE3_BURST	R		Bu	rst rea	d of PA	CE3_	A, PA	CE	3_B & PACE	3_C registe	rs (80	0-bit	frame	e: 8-	-bit c	omr	nan	d + 3	3*24	4-bit data)	
0x3D	PACE3_A	R		PAC	E3_0	DATA[9	:0]			P3_0RFB	P3_0LST		PACE3_1DATA[9:0]						F	P3_1RFB	P3_1LST	
0x3E	PACE3_B	R		PAC	E3_2	DATA[9	:0]			P3_2RFB	P3_2LST		P	ACE	3_31	DATA	۹[9:0	0]		F	P3_3RFB	P3_3LST
0x3F	PACE3_C	R		PAC	E3_4	DATA[9	:0]			P3_4RFB	P3_4LST		P	ACE	3_51	DATA	A[9:0	0]		F	P3_5RFB	P3_5LST
0x40	PACE4_BURST	R		Bu	rst rea	d of PA	CE4_	A, PA	CE4	4_B & PACE	4_C registe	rs (80	0-bit	frame	e: 8-	-bit c	omr	nan	d + 3	3*24	4-bit data)	
0x41	PACE4_A	R		PAC	E4_0	DATA[9	:0]			P4_0RFB	P4_0LST		P	ACE4	1_1	DATA	A[9:0	0]		F	P4_1RFB	P4_1LST
0x42	PACE4_B	R		PAC	E4_2	DATA[9	:0]			P4_2RFB	P4_2LST		P	ACE4	1_31	DATA	A[9:0	0]		F	P4_3RFB	P4_3LST
0x43	PACE4_C	R		PAC	E4_4	DATA[9	:0]			P4_4RFB	P4_4LST	PACE4_5DATA[9:0]						F	P4_5RFB	P4_5LST		
0x44	PACE5_BURST	R		Bu	rst rea	d of PA	CE5_	A, PA	CE	5_B & PACE	5_C registe	rs (80	0-bit	frame	e: 8-	-bit c	omr	nan	d + 3	3*24	4-bit data)	
0x45	PACE5_A	R		PAC	E5_0	DATA[9	:0]	P5_0RFB	P5_0LST	PACE5_1DATA[9:0]							F	P5_1RFB	P5_1LST			
0x46	PACE5_B	R		PAC	E5_2	DATA[9	:0]	P5_2RFB	P5_2LST	PACE5_3DATA[9:0]							F	P5_3RFB	P5_3LST			
0x47	PACE5_C	R		PAC	E5_4	DATA[9	:0]			P5_4RFB	P5_4LST		P	ACE	5_5	DATA	٩[9:0	0]		F	P5_5RFB	P5_5LST

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The pace register groups are written sequentially in time as groups of pace edges are found between ECG data samples starting with PACE0 and written in a circular fashion such that after PACE5 is written then PACE0 will be the next group written. Within each pace group, the data for each pace edge is also written sequentially in time by segment starting with edge 0 but is not written in a circular fashion such that only the first six pace edges between ECG data samples is written to each pace group. If there are more than six edges in a pace group then this data will not be stored and will be lost. A subgroup register written with data for either one or two pace edges is marked as unread and if just the first segment is written then the second segment will be set to 0xFFF. A sub-group register not written with any pace edge data will be set to 0xFFF FFF and marked as read. All unread subgroups need to be read in order for the pace group to be marked as read. A register is marked as read on the 32nd SCLK rising edge in a normal (single word) mode read. There are burst mode registers for each pace register group in order to read all three sub-groups (A, B, and C) during the same serial data transfer. During the burst mode, the sub-groups are marked as read on the 32nd, 56th, and 80th SCLK rising edges for sub-groups A, B, and C, respectively. Burst mode cycles beyond the 80th SCLK edge will not continue read back with the next pace register group; instead the data returned will read 0xFFF.

Whenever a set of pace edges are detected between ECG data samples, the pace Interrupt bit (PINT) is asserted, alerting the µC that there is new pace data ready for read back. The µC should first read back the ECG FIFO data to the point where the PTAG'd samples are identified, and then read back the linked PACE register group, ensuring the pace events are associated with the correct ECG data samples. Examples are provided below. If new pace edge information is written to a previously written and unread PACE register group then the pace overflow status bit, POVR will be asserted and the association with the ECG data sample will be corrupted. In the event that the data is corrupted then either a SYNCH or FIFO RST command should be executed to restore synchronization between the ECG data samples and the PACE register groups.

Table 52. PACE0 to PACE5 (0x30 to 0x47) Register Functionality

INDEX	NAME	DEFAULT	FUNCTION
D[23:14], D[11:2]	PACEx_yDATA[9:0]	0x3FF	Pace Edge Timing Data Pace Edge Timing = PACEx_yDATA[9:0]*t _{RES} where t _{RES} = 1/(2*f _{MSTR}) and is set by the FMSTR[1:0] bits in the CNFG_GEN register. The time is relative to the associated ECG data sample. x = 0 to 5 and is the pace group associated with a specific ECG data output sample. y = 0 to 5 and is the numbered order of the pace edges detected in time.
D[13], D[1]	Px_yRFB	1	Pace Edge Polarity 0 = Falling Edge 1 = Rising Edge x = 0 to 5 and is the pace group associated with a specific ECG data output sample. y = 0 to 5 and is the numbered order of the pace edges detected in time.
D[12], D[0]	Px_yLST	1	Last Pace Edge 0 = Additional pace edges detected in the group 1 = Last pace edge detected in the group or an empty record. x = 0 to 5 and is the pace group associated with a specific ECG data output sample. y = 0 to 5 and is the numbered order of the pace edges detected in time.

ECG and PACE Data Management Examples and Use Cases

The figures and examples below illustrate several valid means of managing an example set of ECG FIFO and PACE register group data. Data for use in the examples is given in the tables below.

 $\underline{\text{Table } 53}$ shows the internal state of the ECG FIFO for purposes of these examples. The example assumes information in locations 0-7 were previously read back (indicated by Y in the READ column) and that data in locations 16 and beyond was either previously read back or empty (indicated by <Y> in the READ column).

Table 53. ECG FIFO Example

										Е	CG F	IFO D	ATA	D[2	3:0]										
READ	INDEX							E	CG_[DATA	[17:0]								ETA	4G[2	:0]	PT	AG[2	:0]
		23 22	2 2	21	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Υ	0		ECG Sample 00 Voltage Data [17:0] = 0x000																-	-	1	1	1	1	
Υ	1		ECG Sample 01 Voltage Data [17:0] = 0x001															-	-	1	1	1	1		
Υ	2		To cample at contage that [o] cheet															1							
Υ	3		200 cample of temage Para [1110] oxoco															1							
Υ	4		ECG Sample 04 Voltage Data [17:0] = 0x004 - 0 0 1 1 1															1							
Υ	5		ECG Sample 05 Voltage Data [17:0] = 0x005															0							
Υ	6		ECG Sample 06 Voltage Data [17:0] = 0x006														1								
Υ	7					ECG	Sam	ple 07	7 Vol	tage	Data	[17:0]	= 0x	:007						-	0	0	1	1	1
	8					ECG	Sam	ple 08	3 Vol	tage	Data	[17:0]	= 0x	800						-	0	0	1	1	1
	9					ECG	Sam	ple 09) Vol	tage	Data	[17:0]	= 0x	009						-	0	0	1	1	1
	10					ECG	Sam	ple 10) Vol	tage	Data	[17:0]	= 0x	00A						-	0	0	0	0	1
	11					ECG	Sam	ple 11	Vol	tage	Data [[17:0]	= 0x	00B						-	0	0	0	1	0
	12					ECG	Sam	ple 12	2 Volt	tage	Data [17:0]	= 0x	00C						-	0	0	1	1	1
	13					ECG	Sam	ple 13	3 Volt	tage	Data [[17:0]	= 0x	00D						-	0	0	1	1	1
	14					ECG	Sam	ple 14	l Vol	tage	Data	[17:0]	= 0x	00E						-	0	0	1	1	1
	15					ECG	Sam	ple 15	5 Vol	tage	Data	[17:0]	= 0x	00F						-	0	0	1	1	1
<y></y>	16								Е	MPT	Y														
<y></y>	17								Ε	MPT	Y														

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Table 54 shows the internal state of the first four groups in the PACE register group for purposes of these examples. The example assumes information in group 0 was previously read back (indicated by Y in the READ column), that

unused words in active groups 1 and 2 were internally marked as read (indicated by <Y> in the READ column), and that the empty groups 3, 4, and 5 are also internally marked as read and filled with default data.

Table 54. PACE FIFO Example

												PAC	E DA	TA D	[23:0]									
READ	INDEX		Edg	ge T	imin	g D	ata	Segi	ment	[9	:0]	RFB	LST		Edge	e Tim	ing	Dat	a S	egm	ent	[9:0]	RFB	LST
		23	22	21	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Υ	0A	Gro	oup (0: E	dge () Tin	ning	Dat	a [9:0)] =	0x000	1	0	Gro	oup 0:	Edg	e 1	Timir	ng I	Data	[9:0]	= 0	x011	0	0
Υ	0B	Gro	oup (0: E	dge 2	2 Tin	ning	Dat	a [9:0)] =	0x022	1	0	Gro	oup 0:	Edg	e 3	Timir	ng [Data	[9:0]	= 0	k033	0	1
Υ	0C	Gro	oup (0: Ed	dge 4	l Tin	ning	Dat	a [9:0)] =	0x3FF	1	1	Gro	up 0:	Edg	e 5	Timir	ng [Data	[9:0]	= 0	κ3FF	1	1
	1A	Gro	oup	1: E	dge () Tin	ning	Dat	a [9:0)] =	0x100	1	0	Gro	up 1:	Edg	e 1	Timir	ng [Data	[9:0]	= 0	k108	0	0
	1B	Gro	oup	1: E	dge 2	2 Tin	ning	Dat	a [9:0)] =	0x110	1	1	Gro	up 1:	Edg	e 3 [·]	Timir	ng [Data	[9:0]	= 0	κ3FF	1	1
<y></y>	1C	Gro	oup	1: Ed	dge 4	l Tin	ning	Dat	a [9:0)] =	0x3FF	1	1	Gro	up 1:	Edg	e 5	Timir	ng [Data	[9:0]	= 0	(3FF	1	1
	2A	Gro	oup 2	2: Ec	dge () Tin	ning	Dat	a [9:0)] =	0x0A0	0	1	Gro	up 2:	Edg	e 1	Timir	ng [Data	[9:0]	= 0	(3FF	1	1
<y></y>	2B	Gro	oup 2	2: E	dge 2	? Tin	ning	Dat	a [9:0)] =	0x3FF	1	1	Gro	up 2:	Edg	e 3 [·]	Timir	ng [Data	[9:0]	= 0:	(3FF	1	1
<y></y>	2C	Gro	oup 2	2: Ec	dge 4	l Tin	ning	Dat	a [9:0)] =	0x3FF	1	1	Gro	up 2:	Edg	e 5	Timir	ng [Data	[9:0]	= 0	(3FF	1	1
<y></y>	3A	Gro	oup (3: E	dge () Tin	ning	Dat	a [9:0)] =	0x3FF	1	1	Gro	up 3:	Edg	e 1 [·]	Timir	ng [Data	[9:0]	= 0	κ3FF	1	1
<y></y>	3B	Gro	oup (3: E	dge 2	? Tin	ning	Dat	a [9:0)] =	0x3FF	1	1	Gro	up 3:	Edg	e 3 [·]	Timir	ng [Data	[9:0]	= 0	(3FF	1	1
<y></y>	3C	Gro	oup (3: E	dge 4	l Tin	ning	Dat	a [9:0)] =	0x3FF	1	1	Gro	up 3:	Edg	e 5	Timir	ng [Data	[9:0]	= 0	(3FF	1	1
<y></y>	4A	Gro	oup 4	4: Ec	dge C) Tin	ning	Dat	a [9:0)] =	0x3FF	1	1	Gro	up 4:	Edg	e 1	Timir	ng [Data	[9:0]	= 0	κ3FF	1	1
<y></y>	4B	Gro	up 4	4: Ed	dge 2	? Tin	ning	Dat	a [9:0)] =	0x3FF	1	1	Gro	up 4:	Edg	e 3	Timir	ng [Data	[9:0]	= 0	(3FF	1	1
<y></y>	4C	Gro	up 4	4: E	dge 4	Tin	ning	Dat	a [9:0)] =	0x3FF	1	1	Gro	up 4:	Edg	e 5	Timir	ng [Data	[9:0]	= 0:	(3FF	1	1
<y></y>	5A	Gro	up (5: E	dge C	Tin	ning	Dat	a [9:0)] =	0x3FF	1	1	Gro	up 5:	Edg	e 1	Timir	ng [Data	[9:0]	= 0:	k3FF	1	1
<y></y>	5B	Gro	up (5: E	dge 2	? Tin	ning	Dat	a [9:0)] =	0x3FF	1	1	Gro	up 5:	Edg	e 3	Timir	ng [Data	[9:0]	= 0	(3FF	1	1
<y></y>	5C	Gro	up :	5: E	dge 4	l Tin	ning	Dat	a [9:0)] =	0x3FF	1	1	Gro	up 5:	Edg	e 5	Timir	ng [Data	[9:0]	= 0:	(3FF	1	1

ECG Interrupt Driven Normal Mode Example

In this example, the μ C reads back ECG and pace data in response to EINT being asserted and interrupting the μ C via INTB or INT2B and that EFIT=8. For the samples given, the following SPI transactions might result:

The example below will read back complete and correct results but better use could be made of the ECG ETAG and pace information to realize more efficient μ C communications.

Table 55. ECG FIFO and PACE Register Read Back Example (EINT, Normal Mode)

													FIF	DAT	4 D[2	3:0]				-						
CMD	FIFO	INDEX	23	22	21	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
								E	CG S	amp	le Vo	Itage	Data	17:0]	,						ET	AG[2	:0]	ı	PTAG[2	:0]
0x21	ECG	8					E	CG S	ampl	e 08	Volta	ge D	ata [17:	0] = 0	k008						0	0	0	1	1	1
0x21	ECG	9					Е	CG S	ampl	e 09	Volta	ge D	ata [17:	0] = 0>	(009						0	0	0	1	1	1
0x21	ECG	10					E	CG S	ampl	e 10	Volta	ge D	ata [17:	0] = 0>	(00A						0	0	0	0	0	1
0x21	ECG	11					E	CG S	ampl	e 11	Volta	ge Da	ata [17:	0] = 0x	00B						0	0	0	0	1	0
0x21	ECG	12					E	CG S	ampl	e 12	Volta	ge Da	ata [17:	0] = 0x	00C						0	0	0	1	1	1
0x21	ECG	13					E	CG S	ampl	e 13	Volta	ge Da	ata [17:	0] = 0x	00D						0	0	0	1	1	1
0x21	ECG	14					E	CG S	ampl	e 14	Volta	ge Da	ata [17:	0] = 0x	00E						0	0	0	1	1	1
0x21	ECG	15					E	CG S	ampl	e 15	Volta	ge D	ata [17:	0] = 0>	(00F						0	1	0	1	1	1
0x21	ECG							ECG	Emp	ty Vo	oltage	Data	a [17:0]	= 0x0	00						1	1	0	1	1	1
			23	22	21	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
				Е	dge	Timin	ıg Da	ta Se	gme	nt [9	:0]		RFB	LST		Edg	је Т	imin	g Da	ata S	Segm	ent [9:0]		RFB	LST
0x35	PACE	1A	G	rou	p 1: E	Edge	0 Timi	ing D	ata [9	9:0] =	0x10	0	1	0	Gr	oup 1	: E	dge 1	1 Tin	ning	Data	[9:0]	= 0x1	80	0	0
0x36	PACE	1B	G	Grou	p 1: E	Edge	2 Tim	ing D	ata [9	9:0] =	0x11	0	1	1	Gr	oup 1	: Ec	lge 3	3 Tin	ning	Data	[9:0] =	= 0x3	FF	1	1
0x37	PACE	1C	G	Group 1: Edge 4 Timing Data [9:0] = 0x3FF 1 1 Group 1: Edge 5 Timing Data									Data	[9:0]	= 0x3	FF	1	1								
0x39	PACE	2A	G	Group 2: Edge 0 Timing Data [9:0] = 0x3FF										1	1											
0x3A	PACE	2B	G	Group 2: Edge 2 Timing Data [9:0] = 0x3FF											FF	1	1									
0x3B	PACE	2C	G	rou	p 2: E	Edge 4	4 Timi	ing Da	ata [9):0] =	0x3F	F	1	1	Gr	oup 2	: Ec	lge 5	5 Tin	ning	Data	[9:0] =	= 0x3	FF	1	1

The example transactions below will read back identical results, but μ C communication efficiency is improved by only reading back necessary locations, as indicated by the ECG ETAG and PACE LST bits.

Table 56. ECG FIFO and PACE Register Read Back Example (EINT, Normal Mode, Reduced Transactions)

														FIFO	DATA	D[23:0	0]										
CMD	FIFO	INDEX	23	22	21	20	19	1	18 1	7 1	6	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
							•		EC	3 Sa	mp	ole V	oltag	je Data	[17:0]							ET	AG[2	2:0]		PTAG[2	2:0]
0x21	ECG	8					Е	CC	G Sar	nple	08	Volt	age [Data [17	:0] = 0x	800						0	0	0	1	1	1
0x21	ECG	9					Е	CC	G Sar	nple	09	Volt	age [Data [17	:0] = 0x	009						0	0	0	1	1	1
0x21	ECG	10					Е	CC	G Sar	nple	10	Volt	age [Data [17	:0] = 0x	00A						0	0	0	0	0	1
0x21	ECG	11					Е	CC	G Sar	nple	11	Volta	age [) Data [17:	0] = 0x	00C						0	0	0	0	1	0
0x21	ECG	12					Е	CC	G San	nple	12	Volta	age [Data [17:	0] = 0x	00D						0	0	0	1	1	1
0x21	ECG	13					Е	CC	G San	nple	13	Volt	age [Data [17	:0] = 0x	00E						0	0	0	1	1	1
0x21	ECG	14					Е	CC	G Sar	nple	14	Volt	age [Data [17	:0] = 0x	00F						0	0	0	1	1	1
0x21	ECG	15					Е	CC	G Sar	nple	15	Volt	age [Data [17	:0] = 0x	00F						0	1	0	1	1	1
			23	22	21	20	19	1	18 17	7 1	6	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
				Edge Timing Data Segment [9:0] RFB LST Edge Timing Data Segment [9:										9:0]		RFB	LST										
0x35	PACE	1A	Gı	Group 1: Edge 0 Timing Data [9:0] = 0x100											9:0] =	= 0x1	108	0	0								
0x36	PACE	1B	G	Group 1: Edge 2 Timing Data [9:0] = 0x100												BFF	1	1									
0x39	PACE	2A	Gı	roup 2	2: Ec	lge 0	Timin	ng l	Data	[9:0]	= (0x0A	0	0	1	Grou	ıp 2: l	Edge	1 Ti	min	g D	ata [9:0] =	= 0x3	BFF	1	1

PACE Interrupt Driven Normal Mode Example

In this example, the μ C reads back data in response to PINT, which will be asserted in response to the two detected pace events (before EINT will be issued since the EFIT=8 threshold is not met). Note the ECG information should still be read first in order to properly locate the pace events in time. For the samples given, the following SPI transactions might result (note: other combinations of ETAGs are possible depending on the state of the ECG FIFO when the PINT interrupts were serviced).

Table 57. ECG FIFO and PACE Register Read Back Example (PINT, Normal Mode)

								-				-	FIFO	DATA	D[23	:0]										
REG	FIFO	INDEX	23	22	21	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
								Е	CG	Samı	ple Vo	ltage	Data [17:0]							ET.	AG[2	2:0]	ı	PTAG[2	2:0]
0x21	ECG	8					Е	CG S	Sam	ple 08	3 Volta	ge Da	ata [17:	0] = 0x	(008						0	0	0	1	1	1
0x21	ECG	9					Е	CG S	Sam	ple 09) Volta	ge Da	ata [17:	0] = 0x	(009						0	0	0	1	1	1
0x21	ECG	10					Е	CG S	Sam	ple 10) Volta	ge Da	ita [17:	0] = 0x	A00						0	0	0	0	0	1
				E	dge	Timir	ng Da	ıta Se	gm	ent [9:0]		RFB	LST		Edge	Tir	ning	Da	ta S	egme	ent [9:0]		RFB	LST
0x35	PACE	1A		Group 1: Edge 0 Timing Data [9:0] = 0x100												108	0	0								
0x36	PACE	1B	(Group	1: E	Edge	2 Tim	ning D	ata	[9:0] =	= 0x11	0	1	1	Gro	up 1:	Edç	je 3	Timi	ing E)ata [9:0]	= 0x3	3FF	1	1
0x37	PACE	1C	(Group	1: E	Edge 4	4 Tim	ing D	ata	[9:0] =	= 0x3F	F	1	1	Gro	up 1:	Edç	je 5	Timi	ing E)ata [9:0]	= 0x3	3FF	1	1
								Е	CG	Samı	ple Vo	ltage	Data [17:0]							ET	AG[2	2:0]	1	PTAG[2	2:0]
0x21	ECG	11					Е	CG S	Sam	ple 11	Volta	ge Da	ta [17:	0] = 0x	00B						0	0	0	0	1	0
				E	dge '	Timir	ng Da	ıta Se	gm	ent [9:0]		RFB	LST		Edge	Tir	ning	Da	ta S	egme	ent [[9:0]		RFB	LST
0x39	PACE	2A	C	Group	2: E	Edge	0 Tim	ing D	ata	[9:0] =	= 0x0 <i>P</i>	۰0	0	1	Gro	up 2:	Edç	je 1	Timi	ing D)ata [9:0]	= 0x3	3FF	1	1
0x3A	PACE	2B	(Group	2: E	Edge :	2 Tim	ing D	ata	[9:0] =	= 0x3F	F	1	1	Gro	up 2:	Edç	je 3	Timi	ing D)ata [9:0]	= 0x3	3FF	1	1
0x3B	PACE	2C	0	Group	2: E	Edge 4	4 Tim	ing D	ata	[9:0] =	= 0x3F	F	1	1	Gro	up 2:	Edg	je 5	Timi	ing D)ata [9:0]	= 0x3	3FF	1	1
								E	CG	Samı	ple Vo	ltage	Data [17:0]							ET.	AG[2	2:0]	ı	PTAG[2	2:0]
0x21	ECG	12		ECG Sample 12 Voltage Data [17:0] = 0x00D											0	0	0	1	1	1						
0x21	ECG	13		ECG Sample 13 Voltage Data [17:0] = 0x00D 0										0	0	0	1	1	1							
0x21	ECG	14		ECG Sample 14 Voltage Data [17:0] = 0x00E											1	1	1									
0x21	ECG	15		ECG Sample 15 Voltage Data [17:0] = 0x00F												1	1	1								
0x21	ECG															1	1	1								

In the example above, the μ C will read back complete and correct results but better use could be made of the ECG ETAG and pace information to realize more efficient μ C communications as shown below.

The example transactions above will read back identical results, but the efficiency is improved by only reading back locations indicated by the ECG ETAG and PACE LST bits.

Table 58. ECG FIFO and PACE Register Read Back Example (PINT, Normal Mode, Reduced Transactions)

													FIFO	DATA I	D[23:	:0]											
REG	FIFO	INDEX	23	22	21	20	19	18	1	7 16	15	14	13	12	11	1	0 9	1	3	7	6	5	4	3	2	1	0
								EC	GS	Sample	Volta	age D	ata [17	7:0]		_						ET	AG[2	:0]	F	PTAG[2	:0]
0x21	ECG	8					EC	G Saı	mpl	le 08 Vo	oltage	Data	a [17:0]	= 0x00	08							0	0	0	1	1	1
0x21	ECG	9					EC	G Saı	mpl	le 09 Vo	oltage	Data	a [17:0]	= 0x00	09							0	0	0	1	1	1
0x21	ECG	10					EC	G Saı	mpl	le 10 Vo	ltage	Data	[17:0]	= 0x00	DΑ							0	0	0	0	0	1
			Edge Timing Data Segment [9:0] RFB LST Edge Timing I											g Da	ata	Segr	nent	[9:0]		RFB	LST						
0x35	PACE	1A	Group 1: Edge 0 Timing Data [9:0] = 0x100											ning	Data	a [9:0] = 0x	108	0	0							
0x36	PACE	1B	G	Group 1: Edge 2 Timing Data [9:0] = 0x110										Tim	ning	Data	a [9:0] = 0x	3FF	1	1						
								EC	G S	Sample	Volta	age D	ata [17	7:0]								ET	AG[2	:0]	F	PTAG[2	:0]
0x21	ECG	11					EC	G Saı	mpl	le 11 Vo	ltage	Data	[17:0]	= 0x00)B							0	0	0	0	1	0
				Ec	lge T	iming	Dat	a Seç	gm	ent [9:	0]		RFB	LST		Е	dge ⁻	Γin	in	g Da	ata	Segr	nent	[9:0]		RFB	LST
0x39	PACE	2A	G	roup	2: E	dge 0	Timir	ng Da	ata	[9:0] = ()x0A)	0	1	Gro	oup	2: E	dg	e 1	Tim	ning	Data	a [9:0] = 0x	3FF	1	1
			ECG Sample Voltage Data [17:0]												ET	AG[2	:0]	F	PTAG[2	::0]							
0x21	ECG	12		ECG Sample 12 Voltage Data [17:0] = 0x00C												0	0	0	1	1	1						
0x21	ECG	13		ECG Sample 13 Voltage Data [17:0] = 0x00D												0	0	0	1	1	1						
0x21	ECG	14		ECG Sample 14 Voltage Data [17:0] = 0x00E												1	1	1									
0x21	ECG	15		ECG Sample 14 Voltage Data [17:0] = 0x00E												1	1										

Burst Mode Example

In this example, the μ C reads data in response to the EINT bit and that EFIT = 8. For the samples given, the following Burst Mode SPI transactions might result.

The example burst mode transactions below will read back complete and correct results. Note that to achieve this read back in burst mode only three commands are issued: ECG Burst 8 + (9×24) SCLK cycles, PACE Group 1 Burst 8 + (3×24) SCLK cycles, and PACE Group 2 Burst 8 + (3×24) SCLK cycles; however, better use could be made of the ECG ETAG and pace information to realize more efficient μ C communications.

Table 59. ECG FIFO and PACE Register Read Back Example (EINT, Burst Mode)

													FIFO	DATA			Ť									
REG	FIFO	INDEX	23	22	24	20	19	4	8 17	16	15	14	13	12	11	_	9	8	7	6	5	4	3	2	1	0
			23	22	21	20	19	10							11	10	9	0	1	ь	5	4	3		1	U
									ECG	Samı	ole Vo	oltag	e Data	17:0]							ET.	AG[2	2:0]		PTAG[2	2:0]
0x20	ECG	8					E	EC	G Samp	ole 08	Volta	age D	ata [17:	0] = 0x	800						0	0	0	1	1	1
	ECG	9					Е	EC	G Samp	ole 09	Volta	age D	ata [17:	0] = 0x	009						0	0	0	1	1	1
	ECG	10					Е	ECC	G Samp	le 10	Volta	ige D	ata [17:	0] = 0x	:00A						0	0	0	0	0	1
	ECG	11					Е	ECC	G Samp	le 11	Volta	ige D	ata [17:	0] = 0x	00B						0	0	0	0	1	0
	ECG	12					Е	ECC	G Samp	le 12	Volta	ige D	ata [17:	0] = 0x	00C						0	0	0	1	1	1
	ECG	13					Е	ECC	G Samp	le 13	Volta	ige D	ata [17:	0] = 0x	00D						0	0	0	1	1	1
	ECG	14					Е	ECC	G Samp	le 14	Volta	ige D	ata [17:	0] = 0x	00E						0	0	0	1	1	1
	ECG	15					Е	ECC	G Samp	le 15	Volta	age D	ata [17:	0] = 0x	:00F						0	1	0	1	1	1
	ECG							Е	CG Em	pty V	oltage	e Dat	a [17:0]	= 0x00	00						1	1	0	1	1	1
			23	22	21	20	19	18	8 17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
				E	dge T	imin	g Da	ta :	Segme	nt [9	:0]		RFB	LST		Edge	Tin	ning	Data	a Se	gme	nt [9:0]		RFB	LST
0x34	PACE	1A	(Group	1: E	dge () Timi	ing	Data [9	9:0] =	0x10	0	1	0	Gro	up 1:	Edg	e 1 7	Γimir	ng D	ata [9:0] =	= 0x1	08	0	0
	PACE	1B	(Group	1: E	dge 2	2 Timi	ing	Data [9:0] =	0x11	0	1	1	Gro	up 1:	Edg	e 3 T	Timir	ng D	ata [9:0] =	= 0x3	FF	1	1
	PACE	1C	(Group	1: E	dge 4	Timi	ing	Data [9	9:0] =	0x3F	F	1	1	Gro	up 1:	Edg	e 5 T	Timir	ng D	ata [9:0] =	= 0x3	FF	1	1
0x38	PACE	2A	(Group 2: Edge 0 Timing Data [9:0] = 0x0A0 0 1 Group 2: Edge 1 Timing Data [9:0] = 0x3FF										FF	1	1										
	PACE	2B	(Group 2: Edge 2 Timing Data [9:0] = 0x3FF											FF	1	1									
	PACE	2C	(Group 2: Edge 4 Timing Data [9:0] = 0x3FF											1	1										

The example burst mode transactions below will read back identical results, but the efficiency is improved by only reading back locations indicated by the ECG ETAG and PACE LST bits. To achieve this read back in burst mode only three commands are issued: ECG Burst 8 + (8 x 24) SCLK cycles, PACE Group 1 Burst 8 + (2 x 24) SCLK cycles, and PACE Group 2 Burst 8 + 24 SCLK cycles.

Table 60. ECG FIFO and PACE Register Read Back Example (EINT, Burst Mode, Reduced Transactions)

DE0	5150	MIDEN											FIFO D	ATA ([0[23:	0])										
REG	FIFO	INDEX	23	22	21	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
								Е	CG S	amı	ole Vo	tage	Data [17:0]							ET	AG[2	2:0]	ı	PTAG[2	2:0]
0x20	ECG	8					Е	CG S	ampl	e 08	Volta	ge Da	ita [17:0	0) = 0x	800						0	0	0	1	1	1
	ECG	9					Е	CG S	ampl	e 09	Volta	ge Da	ta [17:0	0) = 0x	009						0	0	0	1	1	1
	ECG	10					Е	CG S	ampl	e 10	Volta	ge Da	ta [17:0)] = 0x(00A						0	0	0	0	0	1
	ECG	11					Е	CG S	ampl	e 11	Voltag	je Da	ta [17:0)] = 0x(00B						0	0	0	0	1	0
	ECG	12					E	CG S	ample	e 12	Voltag	je Da	ta [17:0)] = 0x(00C						0	0	0	1	1	1
	ECG	13					E	CG S	ample	e 13	Voltag	je Da	ta [17:0)] = 0x(00D						0	0	0	1	1	1
	ECG	14					E	CG S	ampl	e 14	Voltag	ge Da	ta [17:0)] = 0x(00E						0	0	0	1	1	1
	ECG	15					Е	CG S	ampl	e 15	Volta	ge Da	ta [17:0)] = 0x(00F						0	1	0	1	1	1
			23	22	21	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
				Edge Timing Data Segment [9:0] RFB LST Edge Timing Data Segm										gme	ent [9:0]		RFB	LST							
0x34	PACE	1A	(= 0x ²	108	0	0							
	PACE	1B	(Group	1: E	dge 2	2 Timi	ing Da	ata [9	:0] =	= 0x110)	1	1	Gro	up 1:	Edg	e 3	Timiı	ng D	ata [9:0] =	= 0x3	BFF	1	1
0x38	PACE	2A	(Edge Timing Data Segment [9:0]RFBLSTEdge Timing Data Segment [9:0]Group 1: Edge 0 Timing Data [9:0] = 0x10010Group 1: Edge 1 Timing Data [9:0] = 0x108													BFF	1	1							

Resulting Data Record Example

In this example, the μ C reads data in response to EINT and that EFIT=8. For the complete FIFO samples given and the resulting two interrupts, the following SPI transactions might have resulted (starting from the beginning of the FIFO record).

Table 61. Complete Read Back Example (EINT, Normal Mode)

DEO	FIFO	INDEX												FIFC	DATA	(D[2	3:0)])											
REG	FIFO	INDEX	23	22	21	20	19	1	8 17	16	15	; ·	14	13	12	11		10	9	8		7	6	5	4	3	2	1	0
									EC	3 Sa	mple	e Ve	olta	ge Data	[17:0]									ET	AG[2:0]		PTAG[2	2:0]
0x21	ECG	0						EC	G Sar	nple	00 V	olta	age	Data [1	7:0] = 0	000x								0	0	1	1	1	1
0x21	ECG	1						EC	G Sar	nple	01 V	olta	age	Data [1	7:0] = ()x001								0	0	1	1	1	1
0x21	ECG	2						EC	G Sar	nple	02 V	olta	age	Data [1	7:0] = 0)x002								0	0	0	1	1	1
0x21	ECG	3						EC	G Sar	nple	03 V	olta	age	Data [1	7:0] = 0)x003								0	0	0	1	1	1
0x21	ECG	4						EC	G Sar	nple	04 V	olta	age	Data [1	7:0] = 0)x004								0	0	0	1	1	1
0x21	ECG	5						EC	G Sar	nple	05 V	olta	age	Data [1	7:0] = 0)x005								0	0	0	0	0	0
0x21	ECG	6						EC	G Sar	nple	06 V	olta	age	Data [1	7:0] = 0)x006								0	0	0	1	1	1
0x21	ECG	7		ECG Sample 07 Voltage Data [17:0] = 0x007												0	1	0	1	1	1								
0x21	ECG		ECG Empty Voltage Data [17:0] = 0x000										1	1	0	1	1	1											
			Edge Timing Data Segment [9:0] RFB LST Edge Timing Data Seg										gme	nt [9:0]		RFB	LST											
0x31	PACE	0A	Group 0: Edge 0 Timing Data [9:0] = 0x000											ıg D	ata [9:0] =	= 0x0)11	0	0									
0x32	PACE	0B	Gı	roup	0: E	dge 2	Timi	ng	Data [9:0]	= 0x0	022	2	1	0	Gr	ou	p 0: I	Edg	ge 3	Ti	min	g D	ata [9:0] =	= 0x0)33	0	1
0x33	PACE	0C	Gr	roup	0: E	dge 4	Timi	ng	Data [9:0]	= 0x3	3FF	F	1	1	Gro	ou	o 0: E	Edg	je 5	Ti	min	g D	ata [9:0] =	0x3	FF	1	1
									EC	3 Sa	mple	Ve	olta	ge Data	[17:0]									ET	AG[2:0]		PTAG[2	2:0]
0x21	ECG	8						EC	G Sar	nple	08 V	olta	age	Data [1	7:0] = (800x								0	0	0	1	1	1
0x21	ECG	9						EC	G Sar	nple	09 V	olta	age	Data [1	7:0] = 0)x009								0	0	0	1	1	1
0x21	ECG	10						EC	G Sar	nple	10 V	olta	age	Data [1	7:0] = 0	0x00A								0	0	0	0	0	1
0x21	ECG	11						EC	G Sar	nple	11 V	olta	age l	Data [1	7:0] = 0	x00B								0	0	0	0	1	0
0x21	ECG	12					ı	EC	G San	nple	12 V	olta	age	Data [1	7:0] = 0	x00C	;							0	0	0	1	1	1
0x21	ECG	13					I	EC	G San	nple	13 V	olta	age	Data [1	7:0] = 0	x00D)							0	0	0	1	1	1
0x21	ECG	14					-	EC	G San	nple	14 V	olta	age	Data [1	7:0] = 0	x00E								0	0	0	1	1	1
0x21	ECG	15						EC	G Sar	nple	15 V	olta	age	Data [1	7:0] = 0)x00F								0	1	0	1	1	1
0x21	ECG								ECG E	mpty	Volt	ag	e Da	ata [17:	0x0 = 0x0	000								1	1	0	1	1	1
			Edge Timing Data Segment [9:0] RFB LST Edge Timing Data Segment [9:0]											RFB	LST														
0x35	PACE	1A	Group 1: Edge 0 Timing Data [9:0] = 0x100										0	0															
0x36	PACE	1B	Gi	roup	1: E	dge 2	Timi	ing	Data	9:0]	= 0x	110)	1	1	Gro	ou	o 1: E	Edg	je 3	Ti	min	g D	ata [9:0] =	0x3	BFF	1	1
0x37	PACE	1C	Gr	roup	1: E	dge 4	Timi	ng	Data [9:0]	= 0x3	3FF	F	1	1	Gro	ou	o 1: E	Edg	je 5	Ti	min	g D	ata [9:0] =	0x3	BFF	1	1
0x39	PACE	2A	Gr	roup	2: E	dge 0	Timi	ng	Data [9:0]	= 0x0)AC	0	0	1	Gro	ou	o 2: E	Edg	je 1	Ti	min	g D	ata [9:0] =	0x3	SFF	1	1
0x3A	PACE	2B	Gr	roup	2: E	dge 2	Timi	ng	Data [9:0]	= 0x3	3FF	F	1	1	Gro	ou	o 2: E	Edg	je 3	Ti	min	g D	ata [9:0] =	0x3	SFF	1	1
0x3B	PACE	2C	Gr	roup	2: E	dge 4	Timi	ng	Data [9:0]	= 0x3	3FF	F	1	1	Gro	ou	o 2: E	Edg	je 5	Ti	min	g D	ata [9:0] =	0x3	BFF	1	1

MAX30001

Ultra-Low-Power, Single-Channel Integrated Biopotential (ECG, R-to-R, and Pace Detection) and Bioimpedance (BioZ) AFE

The μ C must now prepare a complete record of the ECG waveform given the data observed thus far. All empty samples, which do not represent valid ECG time steps or valid pace edges, will be filtered out. Then the pace edges will be interleaved within the appropriate ECG sample intervals. For purposes of this example, assume

FMSTR[1:0] = 01 and ECG_RATE[1:0] = 10 (in CNFG_GEN and CNFG ECG registers, respectively), thus:

 F_{ECG} = 125sps T_{ECG} = 1/ F_{ECG} = 8ms F_{PACE} = 64,000Hz PACE_RES = 1/ F_{PACE} = 15.625µs

Table 62. Example Post-Processed ECG and PACE Record

TIME (ms)	VOLTAGE (LSBs)	F*	C**	P***	NOTE
0.000	0x000	•			FAST mode engaged – ECG voltage may be invalid
8.000	0x001	•			FAST mode engaged – ECG voltage may be invalid
16.000	0x002				
24.000	0x003				
32.000	0x004				
40.000	0x005		•		Pace edge(s) detected during current sample interval - ECG voltage might be impacted
40.000				1	Pace rising edge detected (0*15.625µs = 0.000ms delayed)
40.266				1	Pace falling edge detected (17*15.625µs = 0.256ms delayed)
40.531				1	Pace rising edge detected (34*15.625µs = 0.531ms delayed)
40.797				1	Pace falling edge detected (51*15.625µs = 0.797ms delayed)
48.000	0x006		•		Pace edge(s) detected during preceding sample interval - ECG voltage might be impacted
56.000	0x007				
64.000	0x008				
72.000	0x009				
80.000	0x00A		•		Pace edge(s) detected during current sample interval - ECG voltage might be impacted
84.000				1	Pace rising edge detected (256*15.625µs = 4.000ms delayed)
84.125				 	Pace falling edge detected (264*15.625µs = 4.125ms delayed)
84.250				1	Pace rising edge detected (272*15.625µs = 4.250ms delayed)
88.000	0x00B		•		Pace edge(s) detected during preceding & current sample interval - ECG voltage might be impacted
90.500				1	Pace falling edge detected (160*15.625µs = 2.500ms delayed)
96.000	0x00C		•		Pace edge(s) detected during preceding sample interval - ECG voltage might be impacted
104.000	0x00D				
112.000	0x00E				
120.000	0x00F				

^{*}F = Fast mode

^{**}C = Sample corrupted by Pace activity

^{***}P = Pace edge

Applications Information

External Filters

External filters are recommended in environments with high levels of EMI to improve noise rejection on the inputs. Select corner frequencies according to the requirements of the channel. The typical application circuits in Figure 17 show examples of input filters, but component values must be modified according to application requirements.

The differential ECG signal occupies frequencies from about 0.05Hz to 200Hz. For applications that require less detail such as fitness monitors, the corner frequency can be lowered to about 40Hz, trading noise immunity for ECG detail. Place the common mode corner frequency about a decade below the AM radio band (535kHz).

The BioZ filter depends on the drive frequency used in the application. Place the differential mode corner frequency several decades higher than the maximum drive frequency. Place the common mode corner frequency higher than the differential mode corner frequency, but lower than the AM radio band.

Body Bias Electrode

Compliance with the common mode input range of the ECG and BioZ channels is achieved by using internal lead bias or by adding a third electrode to drive the body to V_{CM} . The body bias drive electrode improves performance in applications with high electrode impedance or high 50Hz/60Hz coupling. Using V_{CM} drive also improves the input impedance because internal lead bias is disabled.

IEC 60601-2-47 Compliance

IEC 60601-2-47:2012 concerns the basic safety and essential performance of AMBULATORY ELECTROCARDIOGRAPHIC SYSTEMS and the MAX30001 can be used in such systems and be compliant. The MAX30001 has been tested according to the clauses and subclauses that pertain to the analog front end and A/D conversion portions of such systems. With proper system design, a system including the MAX30001 can be certified to the standard.

Typical Application Circuits

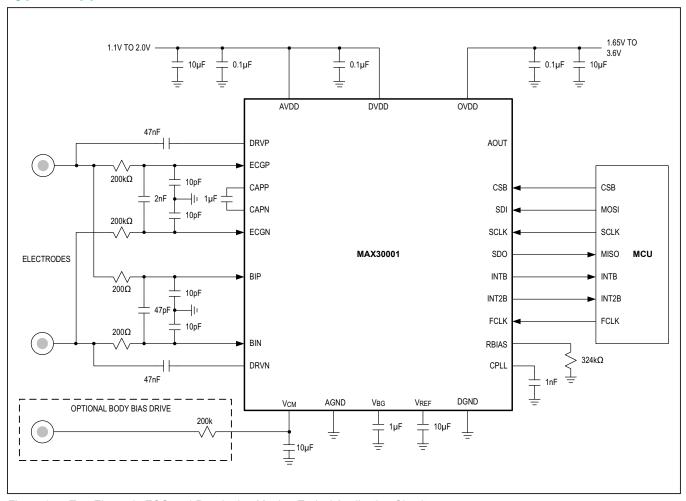


Figure 17a. Two-Electrode ECG and Respiration Monitor Typical Application Circuit

Typical Application Circuits (continued)

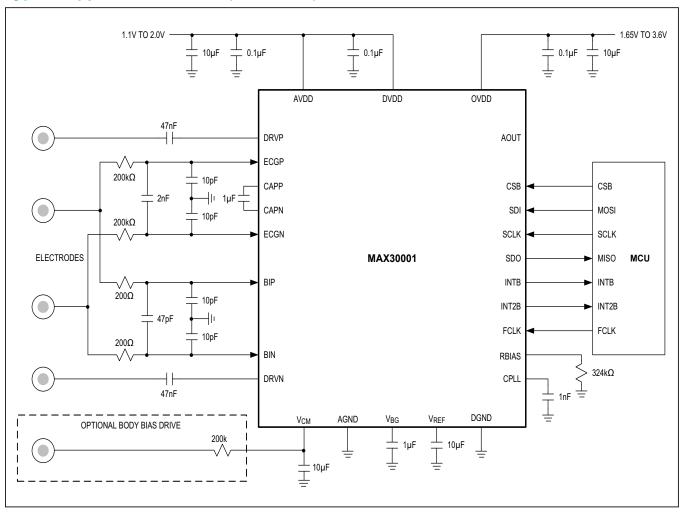


Figure 17b. Four-Electrode ECG and Respiration Monitor Typical Application Circuit

Application Diagrams

See Figure 18 for an example of a clinical application for monitoring ECG and respiration using just two electrodes and with optional shared defibrillation protection circuitry. The electrode models are shown to illustrate the electrical characteristics of the physical electrodes.

Four Electrode ECG and Respiration Monitoring Application

See <u>Figure 19</u> for an example of a clinical application for monitoring ECG and respiration using four electrodes and with optional defibrillation protection circuitry. The electrode models are shown to illustrate the electrical characteristics of the physical electrodes.

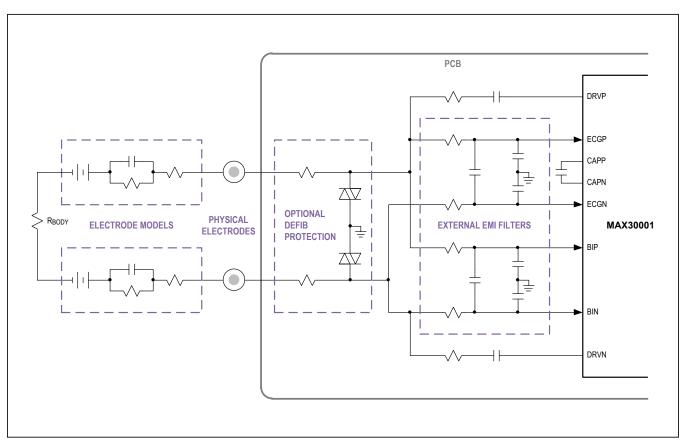


Figure 18. Two Electrode ECG and Respiration Monitoring with Optional Common Defibrillation Protection.

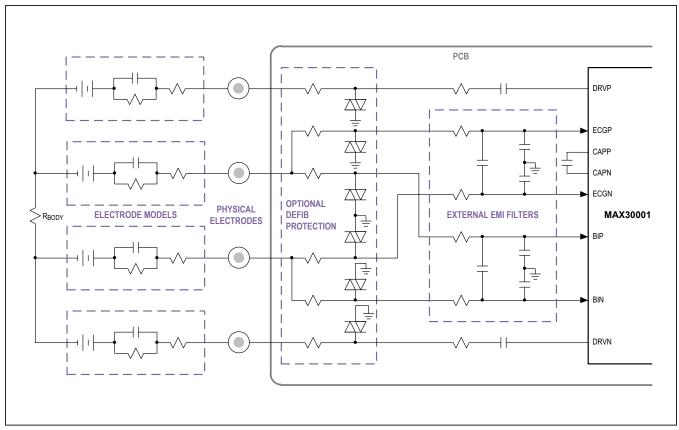


Figure 19. Four Electrode ECG and Respiration Monitoring with Optional Defibrillation Protection.

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Ordering Information

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⁺Denotes lead(Pb)-free/RoHS-compliant package.

Chip Information

PROCESS: CMOS

T = Tape and reel.

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Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	8/17	Initial release	_
1	9/17	Added figures and updated tables	1–86
2	8/19	Updated the General Description, Benefits and Features, Absolute Maximum Ratings, Package Information, Electrical Characteristics, Pin Configuration, Pin Description, ECG Channel, EMI Filtering and ESD Protection, DC Leads-Off Detection and ULP Leads-On Detection, Lead Bias, Gain Settings, Input Range, and Filtering, Fast Recovery, Decimation Filter, BioZ Channel, EMI Filtering and ESD Protection, Leads-Off Detection and ULP Leads-On Detection, Lead Bias, Programmable Resistive Load, Current Generator, Current Selection and Resolution Calculation Example 1 (Two Terminal with Common Protection), Current Selection and Resolution Calculation Example 2 (Four Terminal), Reference and Common Mode Buffer, Table 11, Table 14, Table 19 to Table 21, CNFG_GEN (0x10), Table 32, Table 33, ECG FIFO Data Structure, BioZ FIFO Data Structure, Table 62, and Ordering Information sections; replaced the Functional Diagram, Figure 1a, TOC10-TOC12, TOC17, TOC27, TOC28, TOC34-TOC35, TOC38, Figure 3, Figure 9, Figure 17a, Figure 17b; added the Converting ECG Samples to Voltage, Converting BioZ Samples to Ohms (use symbol), and Application Information section; corrected typos through for subscripting and consistency of symbols	1-8, 10-11, 13-16, 18-22, 24-26, 28-34, 41, 43, 46-47, 53, 65, 67, 80-83, 86

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 AD9826KRSZRL
 AD9860BSTZ
 AD9861BCPZ-50
 AD9862BSTZ
 AD9865BCPZ
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 AD9895KBCZ

 AD9923ABBCZ
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 AD9945KCPZ
 AD9949KCPZ
 AD9963BCPZ
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 ADAS1000-3BSTZ