MAX77818

Dual Input, Power Path, 3A Switch Mode Charger with Fuel Gauge

General Description

The MAX77818 integrates a high-performance 3A switching charger and proprietary ModelGauge™ m5 fuel gauge in a space saving WLP package, ideal for USB powered portable devices. The smart power path charger supports two inputs with reverse blocking and USB On-the-Go (OTG), integrates all power switches, operates at high-switching frequency with high efficiency, enabling low heat designs with small external components. The ModelGauge m5 algorithm combines the excellent short-term accuracy and linearity of a coulomb counter with the long-term stability of a voltage-based fuel gauge, along with temperature compensation to provide industry-leading fuel-gauge accuracy. The device also integrates two high voltage input LDOs and is highly programmable with I²C interface.

Applications

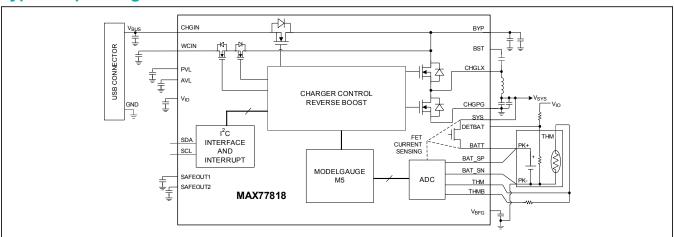
- Industrial PCs
- Portable Medical Devices
- Wearable, Smart Watches
- Handheld Devices
- Wireless Speakers
- Bluetooth Headsets
- Smart Home Automation, Sensors
- Internet of Things (IoT)

Benefits and Features

- Smart Power Path, Dual Input Switching Charger
 - Up to 13.4V Operating/16V Withstand/4A (CHGIN)
 - Up to 6V Operating/1.26A (WCIN)
 - · Reverse Leakage Blocking
 - Reverse Boost with up to 5.1V/1.5A (OTG)
 - Integrated Low-Loss FETs for 3A Charging/4.5A Discharging
 - 95.6% Peak Efficiency at 5V/4.35V
 - · Full Feature Charger with Protection
 - · Thermal Regulation and JEITA Compliance
- ModelGauge m5 Fuel Gauge
 - Time-to-Empty and Time-to-Full Prediction
 - Accurate Battery Capacity (SOC) and Time-to-Empty Readings
 - Temperature, Age, and Discharge Rate Compensated
 - No Calibration
 - Integrated FET Current Sensing
 - Maintains Accuracy without Empty, Full, or Idle States
 - Low Quiescent Current (IQ)
- Two High Voltage Input LDOs
- I²C Serial Interface
- 72-Bump, 3.867mm x 3.608mm, 0.4mm Pitch WLP

Ordering Information appears at end of data sheet.

Typical Operating Circuit



ModelGauge is a trademark of Maxim Integrated Products, Inc.



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Absolute Maximum Ratings

Switching Charger	
CHGIN to GND	0.3V to +16V
BYP to GND	0.3V to +16V
WCIN, PVL, AVL, BAT_SP, BATT, SYS,	
DETBATB to GND	0.3V to +6V
BST to PVL	0.3V to +16V
BST to CHGLX	
WCINOKB, INOKB to GND	0.3V to SYS+0.3V
BAT_SN, CHGPG to GND	
CHGLX, CHGPG Continuous Current	3.5A _{RMS}
SYS, BATT Continuous Current	4.5A _{RMS}
CHGIN, BYP Continuous Current	4.0A _{RMS}
WCIN Continuous Current	1.5A _{RMS}
Fuel Gauge	
V _{BFG} , to GND	0.3V to +2.2V
THMB, THM to GND	$-0.3V$ to $V_{AVL} + 0.3V$

Safeout LDOs SAFEOUT1, SAFEOUT2 to GND SAFEOUT1, SAFEOUT2 Continuous Cu I2C and Interface Logic	
V _{IO} to GND	0.3V to +6V
SDA, SCL to GND	0.3V to V _{IO} +0.3V
INTB to GND	0.3V to $V_{SYS} + 0.3V$
TEST_, V _{CCTEST} , SYS_ to GND	0.3V to +6V
GND_ to GND	
Thermal Ratings	
Operating Temperature Range	40°C to +85°C
Junction Temperature	
Storage Temperature Range	65°C to +150°C
Soldering Temperature (reflow)	+260°C
Continuous Power Dissipation ($T_A = +70$)°C)
(derate 28.9mW/°C with 4L board, abo	ove 70°C)2.31W

CHGLX has internal clamp diodes to CHGPG and BYP. Applications that forward bias these diodes should take care not to exceed the IC's package power dissipation limits.

Package Thermal Characteristics (Note 1)

WLP

Junction-to-Ambient Thermal Resistance (θ_{JA})34.6°C/W

Note 1: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to www.maximintegrated.com/thermal-tutorial.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Electrical Characteristics

General Electrical Characteristics

 $(V_{SYS} = +3.7V, V_{CHGIN} = 0V, V_{IO} = 1.8V, T_A = -40^{\circ}C$ to $+85^{\circ}C$, unless otherwise noted. Limits are 100% production tested at $T_A = +25^{\circ}C$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Typical values are not guaranteed.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
Chutdown Cupply Current (DATT)		All circuits off		23	50	μА	
Shutdown Supply Current (BATT)		V _{BATT} = 3.6V					
No Lood Cumply Current (DATT)		Fuel gauge is on		5 0	100		
No Load Supply Current (BATT)		All other circuits off, V _{BATT} = 3.6V		50		μA	
SYS INPUT RANGE	SYS INPUT RANGE						
SYS Operating Voltage	V _{SYS}	Guaranteed by V _{SYSUVLO} and V _{SYSOVLO}	2.8		5	V	
SYS Undervoltage Lockout Threshold	V _{SYS_UVLO}	V _{SYS} falling, 200mV hysteresis	2.45	2.5	2.55	V	
SYS Overvoltage Lockout Threshold	V _{SYS_OVLO}	V _{SYS} rising, 200mV hysteresis	5.2	5.36	5.52	V	

Switching Charger Electrical Characteristics

 $(V_{CHGIN} = 5V, V_{BATT} = 4.2V, T_A = -40^{\circ}C$ to +85°C, unless otherwise noted. Typical values are at $T_A = 25^{\circ}C$. Fast-charge current is set for 1.5A, done current is set for 150mA. Limits are 100% production tested at $T_A = +25^{\circ}C$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Typical values are not guaranteed.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
CHGIN INPUT	•					
CHGIN Operating Voltage Range		Operating voltage	3.2		V _{OVLO}	V
WCIN Operating Voltage Range		Operating voltage	3.2		V _{OVLO}	V
CHGIN Overvoltage Threshold (Note 4)	V _{CHGIN-OVLO}	V _{CHGIN} rising	13.4	13.7	14	V
WCIN Overvoltage Threshold (Note 4)	V _{WCIN-OVLO}	V _{WCIN} rising	5.9		6	V
WCIN Overvoltage Threshold Hysteresis	Vwcinh-ovlo	V _{WCIN} falling		100		mV
CHGIN Overvoltage Threshold Hysteresis	V _{CHGINH-OVLO}	V _{CHGIN} falling		300		mV
WOIN/CHOIN Overvolters Delay	_	V _{WCIN/BUS_DET} rising, 100mV overdrive, not production tested		10		us
WCIN/CHGIN Overvoltage Delay	T _{D-OVLO}	V _{WCIN/BUS_DET} falling, 100mV overdrive, not production tested		20		us
WCIN/CHGIN to GND Minimum Turn-On Threshold Range (Note 4)	Vwcin_uvlo/ Vchgin_uvlo	V _{CHGIN} rising, 100mV hysteresis, programmable at 4.5V, 4.9V, 5.0V, 5.1V, WCIN input is disabled when valid CHGIN input is detected	4.5		5.1	V
WCIN/CHGIN to GND Minimum Turn-On Threshold Accuracy	VWCIN_UVLO/ VCHGIN_UVLO	V _{WCIN/CHGIN} rising, 4.5V setting	4.4	4.5	4.6	V
WCIN/CHGIN to SYS Minimum Turn-On Threshold (Note 4)	Vwcin2sys/ Vchgin2sys	V _{CHGIN} rising, 50mV hysteresis, WCIN input is disabled when valid CHGIN input is detected	V _{SYS} + 0.12	V _{SYS} + 0.20	V _{SYS} + 0.28	V
WCIN/CHGIN Turn-On Threshold Delay	T _{D-UVLO}	Not production tested		10		us
WCIN/CHGIN Adaptive Current Regulation Threshold Range (Note 5)	Vwcin_reg/ Vchgin_reg	Programmable at 4.3V, 4.7V, 4.8V, 4.9V	4.3		4.9	V
WCIN/CHGIN Adaptive Voltage Regulation Threshold Accuracy	V _{WCIN_REG} / V _{CHGIN_REG}	4.9V setting	4.8	4.9	5	V
CHGIN Current-Limit Range		Programmable, 500mA default, factory programmable option of 100mA, production tested at 100mA, 500mA, 1000mA, 1800mA, 4000mA settings only	0.1		4	А
WCIN Current-Limit Range		Programmable, 500mA default, factory programmable option of 100mA, production tested at 100mA, 250mA, 500mA, 1000mA settings only	0.06		1.26	А

 $(V_{CHGIN} = 5V, V_{BATT} = 4.2V, T_A = -40^{\circ}C$ to +85°C, unless otherwise noted. Typical values are at $T_A = 25^{\circ}C$. Fast-charge current is set for 1.5A, done current is set for 150mA. Limits are 100% production tested at $T_A = +25^{\circ}C$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Typical values are not guaranteed.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
		V _{WCIN/CHGIN} = 2.4V, the input is undervoltage and R _{INSD} is the only loading		0.075		- mA
WCIN or CHGIN Supply Current	I _{IN}	V _{WCIN/CHGIN} = 5.0V, charger disabled		0.17	0.5	
TOTAL OF CITCHE Supply Surferit	1111	$V_{WCIN/CHGIN}$ = 5.0V, charger enabled, V_{SYS} = V_{BATT} = 4.5V, (no switching, battery charged)		2.7	4	
		V_{WCIN} or V_{CHGIN} = 5.0V, charger enabled, V_{BATT} = 3.8V, 100mA input current setting, T_A = +25°C	90	102	108	- mA
VWCIN or VCHGIN Input Current	I	V_{WCIN} or V_{CHGIN} = 5.0V, charger enabled, V_{BATT} = 3.8V, 500mA Input current setting, T_A = +25°C	462.5	487.5	500	
Limit	^I INLIMIT	V_{WCIN} or V_{CHGIN} = 5.0V, charger enabled, V_{BATT} = 3.8V, 1000mA Input current setting, T_A = +25°C	950	975	1000	
		V_{WCIN} or V_{CHGIN} = 5.0V, charger enabled, V_{BATT} = 3.8V, 1000mA input current setting, T_A = 0°C to +85°C	926	975	1024	
	IINLIMIT	V _{CHGIN} = 5.0V, charger enabled, V _{BATT} = 3.8V, 1800mA input current setting, T _A = +25°C	1710	1755	1800	- mA
VCHCIN Input Current Limit		V _{CHGIN} = 5.0V, charger enabled, V _{BATT} = 3.8V, 1800mA input current setting, T _A = 0°C to +85°C	1667	1755	1843	
VCHGIN Input Current Limit		V_{CHGIN} = 5.0V, charger enabled, V_{BATT} = 3.8V, 4000mA input current setting, T_A = +25°C	3800	3900	4000	
		V_{CHGIN} = 5.0V, charger enabled, V_{BATT} = 3.8V, 4000mA input current setting, T_A = 0°C to +85°C	3705	3900	4095	
WCIN, CHGIN Self-Discharge Down to UVLO Time	tinsd	Time required for the charger input to cause a $10\mu F$ input capacitor to decay from 6.0V to 4.3V.		100		ms
WCIN, CHGIN Input Self-Discharge Resistance	R _{INSD}	For CHGIN, this resistor is disconnected from the CHGIN pin during MUIC microphone mode		35		kΩ
WCINOK/CHGINOK to Start Switching	t _{START}			150		ms

 $(V_{CHGIN} = 5V, V_{BATT} = 4.2V, T_A = -40^{\circ}C$ to +85°C, unless otherwise noted. Typical values are at $T_A = 25^{\circ}C$. Fast-charge current is set for 1.5A, done current is set for 150mA. Limits are 100% production tested at $T_A = +25^{\circ}C$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Typical values are not guaranteed.)

PARAMETER	SYMBOL	CONDITION	S	MIN	TYP	MAX	UNITS		
SWITCH IMPEDANCES AND LEA	KAGE CURREN	TS							
CHGIN to BYP Resistance	R _{IN2BYP}	Bidirectional			0.0144	0.04	Ω		
WCIN to BYP Resistance	R _{WCIN2BYP}				0.093	0.26	Ω		
CHGLX High-Side Resistance	R _{HS}				0.0327	0.1	Ω		
CHGLX Low-Side Resistance	R _{LS}				0.0543	0.14	Ω		
BATT to SYS Dropout Resistance	R _{BAT2SYS}				0.0128	0.04	Ω		
CHGIN to BATT Dropout	Pinopat	Calculation estimates a 0.04 resistance (R _L)	IΩ inductor	0.0000		0.0999			Ω
Resistance	R _{IN2BAT}	R _{IN2BAT} = R _{IN2BYP} + R _{HS} R _{BAT2SYS}	+R _L +		0.0999		12		
CHGLX Leakage Current		CHGLX = CHGPG or BYP	T _A = +25°C		0.01	10	μA		
CHOLA Leakage Guilent		CHOLX - CHOPG OF BTP	T _A = +85°C		1		μA		
BST Leakage Current		V _{BST} = 5.5V	T _A = +25°C		0.01	10	μA		
DOT Leakage Current		VBST - 5.5V	T _A = +85°C		1		μA		
DVD Lookogo Current		$V_{BYP} = 5.5V, V_{CHGIN} =$	T _A = +25°C		0.01	10	μA		
BYP Leakage Current		0V, V _{CHGLX} = 0V, charger disabled	T _A = +85°C		1		μA		
WONL I O		$V_{BYP} = 0V$	T _A = +25°C		0.01		μA		
WCIN Leakage Current		V _{CHGIN} = 0V, V _{WCIN} = 5.5V	T _A = +85°C		1		μA		
SVS Lookago Current		$V_{SYS} = 0V$,	T _A = +25°C		0.01	10	μA		
SYS Leakage Current		V _{BATT} = 4.2V, charger disabled	T _A = +85°C		1		μA		

 $(V_{CHGIN} = 5V, V_{BATT} = 4.2V, T_A = -40^{\circ}C$ to +85°C, unless otherwise noted. Typical values are at $T_A = 25^{\circ}C$. Fast-charge current is set for 1.5A, done current is set for 150mA. Limits are 100% production tested at $T_A = +25^{\circ}C$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Typical values are not guaranteed.)

PARAMETER	SYMBOL	CONDITION	S	MIN	TYP	MAX	UNITS
		V _{CHGIN} = 0V, V _{SYS} = 0V,	T _A = +25°C		20	30	μA
		V_{BATT} = 4.2V, QBAT is off	T _A = +85°C		20		μA
		V _{CHGIN} = 0V, V _{BATT} = 4.2V, QBAT is on,	T _A = +25°C		15.3		μA
	I _{MBAT}	main-battery overcurrent protection disabled	T _A = +85°C		15.3		μA
	IMBAT	$V_{BATT} = 4.2V$, QBAT is on,	T _A = +25°C		20		μA
BATT Quiescent Current (I _{SYS} = 0A, I _{BYP} = 0A)		main-battery overcurrent protection enabled	T _A = +85°C		20		μA
		$V_{SYS} = 4.2V$, $V_{BATT} = 0V$,	T _A = +25°C		0.01	10	μA
		charger disabled	T _A = +85°C		1		μA
	la construction of the con	V _{CHGIN} = 5V, V _{BATT} = 4.2V, QBAT is off, main-battery overcurrent	T _A = +25°C		3	10	μA
	I _{MBDN}	protection disabled, Charger is enabled but in its done mode	T _A = +85°C		3		μА
CHARGER DC-DC BUCK							
Minimum On-Time	ton-min				75		ns
Minimum Off-Time	t _{OFF}				75		ns
		$T_A = 0$ °C to +85°C	I _{LIM} = 00 (3.00A out)	4.15	5.05	5.95	
		$I_{ND} = 0$ (0.47µH inductor option)	I _{LIM} = 01 (2.75A out)		4.75		A
		Production tested at I _{LIM} = 00 setting	I _{LIM} = 10 (2.50A out)		4.45		
Current Limit	lene	(Note 7)	I _{LIM} = 11 (2.25A out)		4.15		
(Note 6)	I _{LIM}	$T_A = 0$ °C to +85°C	I _{LIM} = 00 (3.00A out)		4.60		
		I _{ND} = 1 (1.0μH inductor option)	I _{LIM} = 01 (2.75A out)		4.30		A
		Production tested at I _{LIM} = 11 setting	I _{LIM} = 10 (2.50A out)		4.00		
		(Note 7)	I _{LIM} = 11 (2.25A out)	3.00	3.70	4.40	

 $(V_{CHGIN} = 5V, V_{BATT} = 4.2V, T_A = -40^{\circ}C$ to +85°C, unless otherwise noted. Typical values are at $T_A = 25^{\circ}C$. Fast-charge current is set for 1.5A, done current is set for 150mA. Limits are 100% production tested at $T_A = +25^{\circ}C$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Typical values are not guaranteed.)

PARAMETER	SYMBOL	CONDITIO	ONS	MIN	TYP	MAX	UNITS
REVERSE BOOST							
BYP Voltage Adjustment Range		2.5V < V _{BATT} < 4.5V. Ad to 5.75V, production teste 5.75V settings		3		5.75	V
Reverse Boost Quiescent Current	I _{BYP}	Not switching: output force its target regulation voltage			1150		μA
Reverse Boost BYP Voltage in OTG Mode	V _{BYP.OTG}	5.1V setting		4.94	5.1	5.26	V
CHGIN Voltage in OTG Mode	V _{CHGIN.OTG}	Mode = 0x05 or 0x0F, WCIN switch is on, VCHGIN_REG = 4.9V, RIN2WCIN + RDSCHGIN < 300ml, OTG load current ≤ 450mA		4.75			V
			OTG_ILIM = 00	500		550	mA
		3.4V < V _{BATT}	OTG_ILIM = 01	900		990	mA
CHGIN Output Current Limit	ICHGIN.OTG.LIM	< 4.5V, T _A = +25°C	OTG_ILIM = 10	1200		1320	mA
			OTG_ILIM=11	1500		1650	mA
Reverse Boost Output Voltage		Discontinuous inductor current (i.e., skip mode)			±150		mV
Ripple		Continuous inductor curr	ent		±150		mV
CHARGER							
BATT Regulation Voltage Range	V _{BATREG}	Programmable in 25mV sproduction tested at 3.65		3.65		4.7	V
			T _A = +25°C	-0.75		+0.75	%
BATT Regulation Voltage Accuracy		3.65V and 4.7V settings	T _A = 0°C to +85°C	-1		+1	%
Fast-Charge Current Program Range		0A to 3.0A in 50mA steps at 500,1000, 2000 and 30		0		3	А
		Programmed currents ≥ 500mA, V _{BATT} > V _{SYSMIN} (short mode), production tested	T _A = +25°C	-2.5		+2.5	%
Fast-Charge Current Accuracy		at 500mA, 800mA, 1000mA, 2000mA, 3000mA settings	T _A = 0°C to +85°C	-5		+5	%
		Programmed currents ≥ 9 V _{SYSMIN} (LDO mode), p 800mA		-10		+10	%

 $(V_{CHGIN} = 5V, V_{BATT} = 4.2V, T_A = -40^{\circ}C$ to +85°C, unless otherwise noted. Typical values are at $T_A = 25^{\circ}C$. Fast-charge current is set for 1.5A, done current is set for 150mA. Limits are 100% production tested at $T_A = +25^{\circ}C$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Typical values are not guaranteed.)

PARAMETER	SYMBOL	CONI	DITIONS	MIN	TYP	MAX	UNITS
			Programmed for 3.0A	2925	3000	3075	mA
Fast-Charge Currents	I	$T_A = +25$ °C, V_{BATT}	Programmed for 2.0A	1950	2000	2050	mA
rast-onarge ourrents	l _{FC}	> V _{SYSMIN}	Programmed for 1.0A	975	1000	1025	mA
			Programmed for 0.5A	487.5	500	512.5	mA
Low-Battery Prequalification Threshold	V_{PQLB}	V _{BATT} rising		2.8	2.9	3	V
Dead-Battery Prequalification Threshold	V_{PQDB}	V _{BATT} rising		1.9	2	2.1	V
Prequalification Threshold Hysteresis	V _{PQ-H}	Applies to both V _{PQI}	Applies to both V _{PQLB} and V _{PQDB}		100		mV
Low-Battery Prequalification Charge Current	I _{PQLB}	Default setting = disabled		75	100	140	mA
Dead-Battery Prequalification Charge Current	I _{PQDB}			40	55	80	mA
Charger Restart Threshold Range	V _{RSTRT}	Adjustable, 100, 150 be disabled	, and 200; it can also	100	150	200	mV
Charger Restart Deglitch Time		10mV overdrive, 100	ns rise time		130		ms
Top-Off Current Program Range		Programmable from steps.	100 to 350mA in 8	100		350	mA
Top-Off Current Accuracy		Gain				5	%
(Note 8)		Offset				20	mA
Charge Termination Deglitch Time	t _{TERM}	2mV overdrive, 100ns rise/fall time			30		ms
Charger State Change Interrupt Deglitch Time	^t scidg	Excludes transition to timer fault state, watchdog timer state			30		ms
Charger Soft-Start Time	t _{SS}				1.5		ms

 $(V_{CHGIN} = 5V, V_{BATT} = 4.2V, T_A = -40^{\circ}C$ to +85°C, unless otherwise noted. Typical values are at $T_A = 25^{\circ}C$. Fast-charge current is set for 1.5A, done current is set for 150mA. Limits are 100% production tested at $T_A = +25^{\circ}C$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Typical values are not guaranteed.)

PARAMETER	SYMBOL	COND	ITIONS	MIN	TYP	MAX	UNITS
SMART POWER SELECTOR							
		I _{BATT} = 10mA			30		mV
BATT to SYS Reverse Regulation	V _{BSREG}	I _{BATT} = 1A			60		mV
Voltage	BOREG	Load regulation during regulation mode	g the reverse		30		mV/A
Minimum SYS Voltage Accuracy	V _{SYSMIN}	Programmable from 3.4V to 3.7V in 100mV steps, V _{BATT} = 2.8V, tested at 3.4V and 3.7V settings		-3		3	%
Maximum SYS Voltage V _{SYSM}	V	The maximum system voltage: VSYSMAX = VBATREG + RBAT2SYS X IBATT	V _{BATREG} = 4.2V, I _{BATT} = 3.0A		4.245	4.32	V
	V SYSMAX	The maximum system voltage: VSYSMAX = VBATREG + RBAT2SYS x IBATT.	V _{BATREG} = 4.7V, I _{BATT} = 3.0A		4.745	4.82	V
WATCHDOG TIMER							
Watchdog Timer Period	t _{WD}			80			s
Watchdog Timer Accuracy				-20	0	+20	%
CHARGE TIMER							
Prequalification Time	t _{PQ}	Applies to both low-ba			35		min
Fast-Charge Constant Current + Fast-Charge Constant Voltage Time	t _{FC}	Adjustable from 4hrs steps including a disa			8		hrs
Top-Off Time	t _{TO}	Adjustable from 0min steps	to 70min in 10min		30		min
Timer Accuracy				-20		+20	%
AVL FILTER							
Internal AVL Filter Resistance					12.5		Ω

 $(V_{CHGIN} = 5V, V_{BATT} = 4.2V, T_A = -40^{\circ}C$ to +85°C, unless otherwise noted. Typical values are at $T_A = 25^{\circ}C$. Fast-charge current is set for 1.5A, done current is set for 150mA. Limits are 100% production tested at $T_A = +25^{\circ}C$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Typical values are not guaranteed.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
THERMAL FOLDBACK	•					'
Junction Temperature Thermal Regulation Loop Setpoint Program Range	T _{JREG}	Junction temperature when charge current is reduced. Programmable from +85°C to +130°C in 15°C steps, default value is +100°C	85		130	°C
Thermal Regulation Gain	A _{TJREG}	The charge current is decreased 6.7% of the fast charge current setting for every degree that the junction temperature exceeds the thermal regulation temperature. This slope ensures that the full-scale current of 3.0A is reduced to 0A by the time the junction temperature is 20°C above the programmed loop set point. For lower programmed charge currents such as 500mA, this slope is valid for charge current reductions down to 100mA; below 100mA the slope becomes shallower but the charge current still reduced to 0A if the junction temperature is 20°C above the programmed loop set point.		-150		mA/°C
BATTERY OVERCURRENT PRO	TECTION					_
Battery Overcurrent Threshold Range	I _{BOVCR}	Programmable from 3.0A to 4.5A in 0.25A steps, can be disabled	3		4.5	A
Battery Overcurrent Debounce Time	t _{BOVRC}	This is the response time for generating the overcurrent interrupt flag	3	6	10	ms
Battery Overcurrent Protection Quiescent Current	IBOVRC			3 + I _{BATT} /22000		μΑ
System Power-Up Current	I _{SYSPU}		35	50	80	mA
System Power-Up Voltage	V _{SYSPU}	V _{SYS} rising, 100mV hysteresis	2	2.1	2.2	V
System Power-Up Response Time	tsyspu	Time required for circuit to activate from an unpowered state (i.e., main-battery hot insertion)		1		μs
SYSTEM SELF DISCHARGE WIT	TH NO POWER					
BATT Self-Discharge Resistor				600		Ω
SYS Self-Discharge Resistor				600		Ω
Self-Discharge Latch Time				300		ms

 $(V_{CHGIN} = 5V, V_{BATT} = 4.2V, T_A = -40^{\circ}C$ to +85°C, unless otherwise noted. Typical values are at $T_A = 25^{\circ}C$. Fast-charge current is set for 1.5A, done current is set for 150mA. Limits are 100% production tested at $T_A = +25^{\circ}C$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Typical values are not guaranteed.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
DETBATB, INOKB, WCINOKB							
DETBATB Logic Threshold	V _{IH}	4% hysterisis			0.8 x V _{IO}		V
Logic Input Leakage Current	I _{DETBATB}				0.1	1	μA
Output Low Voltage INOKB, WCINOKB		I _{SINK} = 1mA				0.4	V
Output High Leakage INOKB,		\/ - F F\/	T _A = +25°C	-1	0	+1	μA
WCINOKB		$V_{SYS} = 5.5V$	T _A = +85°C		0.1		μA

Safeout LDOs Electrical Characteristics

 $(V_{SYS} = 2.8 \text{V to } 4.5 \text{V}, T_A = -40 ^{\circ}\text{C} \text{ to } 85 ^{\circ}\text{C}, \text{ typical values are at } T_A = +25 ^{\circ}\text{C}, \text{ unless otherwise noted. Limits are } 100 ^{\circ}\text{m} \text{ production tested at } T_A = +25 ^{\circ}\text{C}$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Typical values are not guaranteed.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
SAFEOUT1						
		5V < V _{CHGIN} < 5.5V, I _{OUT} = 10mA, SAFEOUT1 = 0x01 (default)	4.8	4.9	5	V
Output Voltage (Default On)		SAFEOUT1 = 0x00		4.85		V
		SAFEOUT1 = 0x10		4.95		V
		SAFEOUT1 = 0x11		3.3		V
Maximum Output Current		When LDO output is regulated	60			mA
Output Current Limit		When LDO output may be out of regulation	60	150	320	mA
Dropout Voltage		V _{CHGIN} = 5V, I _{OUT} = 60mA		120		mV
Load Regulation		V _{CHGIN} = 5.5V, 30μA < I _{OUT} < 30mA		50		mV
Quiescent Supply Current		Not production tested		72		μA
Output Capacitor for Stable Operation (Note 9)		0μ A < I_{OUT} < 30 mA, MAX ESR = 50 mΩ		1		μF
Minimum Output Capacitor for Stable Operation (Note 9)		0μ A < I_{OUT} < 30 mA, MAX ESR = 50 mΩ		0.7		μF
Internal Off-Discharge Resistance				1200		Ω

Safeout LDOs Electrical Characteristics (continued)

 $(V_{SYS} = 2.8V \text{ to } 4.5V, T_A = -40^{\circ}\text{C} \text{ to } 85^{\circ}\text{C}, \text{ typical values are at } T_A = +25^{\circ}\text{C}, \text{ unless otherwise noted. Limits are } 100\% \text{ production tested at } T_A = +25^{\circ}\text{C}$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Typical values are not guaranteed.))

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
SAFEOUT2						
		5V < V _{CHGIN} < 5.5V, I _{OUT} = 10mA, SAFEOUT2 = 0x01 (default)	4.8	4.9	5	V
Output Voltage (Default Off)		SAFEOUT2 = 0x00		4.85		V
		SAFEOUT2 = 0x10		4.95		V
		SAFEOUT2 = 0x11		3.3		V
Maximum Output Current		When LDO output is regulated	60			mA
Output Current Limit		When LDO output may be out of regulation	60	150	320	mA
Dropout Voltage		V _{CHGIN} = 5V, I _{OUT} = 60mA		120		mV
Load Regulation		V _{CHGIN} = 5.5V, 30μA < I _{OUT} < 30mA		50		mV
Quiescent Supply Current		Not production tested		72		μA
Output Capacitor for Stable Operation (Note 9)		0μ A < I_{OUT} < 30 mA, MAX ESR = 50 mΩ		1		μF
Minimum Output Capacitor for Stable Operation (Note 9)		$0FA < I_{OUT} < 30mA$, MAX ESR = $50mΩ$		0.7		μF
Internal Off-Discharge Resistance				1200		Ω

Fuel Gauge Electrical Characteristics

 $(V_{SYS} = 2.8 \text{V to } 4.5 \text{V}, T_A = -40 ^{\circ}\text{C} \text{ to } 85 ^{\circ}\text{C}, \text{ typical values are at } T_A = +25 ^{\circ}\text{C}, \text{ unless otherwise noted. Limits are } 100\% \text{ production tested at } T_A = +25 ^{\circ}\text{C}$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Typical values are not guaranteed.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
	I _{DD0}	Fuel gauge shut down (Note 10)		0.5		μA
Supply Current	I _{DD1}	Fuel gauge active, average with 7.5% ADC duty cycle (Note 10)		35	70	μA
ADC Duty Cycle	Duty			7.5		%
Parameter Capture Rate	t _{ACQ}	Period of ADC activation loop		0.1758		S
Regulator Output	V_{BFG}		1.5	1.8	1.98	V
VOLTAGE CHANNEL						
V Magaurament Error	V	V _{BATT} = 2.8V to 4.5V, T _A = +25°C	-7.5		+7.5	mV
V _{BATT} Measurement Error	V _{GERR}	$T_A = -40^{\circ}C \text{ to } +85^{\circ}C$	-20		+20	mV
V _{BATT} Measurement Resolution	V _{LSB}			1.25		mV
V _{BATT} Measurement Range	V _{RANGE}		2.8		4.98	V

Fuel Gauge Electrical Characteristics (continued)

 $(V_{SYS} = 2.8 \text{V to } 4.5 \text{V}, T_A = -40 ^{\circ}\text{C} \text{ to } 85 ^{\circ}\text{C}, \text{ typical values are at } T_A = +25 ^{\circ}\text{C}, \text{ unless otherwise noted. Limits are } 100\% \text{ production tested at } T_A = +25 ^{\circ}\text{C}$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Typical values are not guaranteed.)

PARAMETER	SYMBOL	CO	ONDITIONS	MIN	TYP	MAX	UNITS
CURRENT CHANNEL				-			
Current Measurement Resolution	I _{LSB}				1.25		mA
Current Measurement Range	I _{RANGE}			-3.6		+3.6	Α
Current Measurement Offset	I _{OERR}	Long term average at zero input current			±0.25		mA
Current Measurement Symmetrical Error	I _{SERR}	(Notes 11, 12,	(Notes 11, 12, 13)		2%		%
		±3000mA		-150		+150	
Current Measurement Asymmetrical Error	I _{AERR}	±1000mA	(Notes 12, 13, 14)	-20		+20	mA
7.6ymmetrical Error		±300mA		-9.5		+9.5	1
Linear Regulator Mode Current		+1500mA	(Note 15)	-225		+225	0
Measurement Error	ILRERR	+100mA			+40	mA	
Time Dage Assurage		V _{SYS} = 3.7V a	at T _A = +25°C		±1		- %
Time-Base Accuracy	t _{ERR}	$T_A = -40^{\circ}C$ to	+85°C	-3.5		+3.5	70
THERMAL CHANNEL							
Ratiometric Measurement Accuracy, THM	T _{GERR}	(Note 13)		-0.5		+0.5	% of full scale
Ratiometric Measurement Resolution, THM	T _{LSB}				0.0244		% of full scale
THMB Output Drive	V _{OH_THMB}	I _{OH_THMB} = -	0.5mA	V _{AVL} - 0.1			V
THMB Precharge Time	tpre_thmb				12.7		ms
THMB Operating Range	V_{THMB}			2.8		V _{AVL}	V
THMB Input Leakage	I _{IN_THMB}	V _{THMB} = 5V		-1		+1	μA
THM Input leakage	I _{IN_THM}			-1		+1	μA

I²C Electrical Characteristics

 $(V_{SYS} = +3.7V, V_{CHGIN} = 0V, V_{IO} = 1.8V, T_A = -40^{\circ}C$ to $+85^{\circ}C$, unless otherwise noted. Limits are 100% production tested at $T_A = +25^{\circ}C$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Typical values are not guaranteed.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
LOGIC AND CONTROL INPUT	,					,
SCL, SDA Input Low Level		T _A = +25°C			0.3 x V _{IO}	V
SCL, SDA Input High Level		T _A = +25°C	0.7 x V _{IO}			V
SCL, SDA Input Hysteresis		T _A = +25°C		0.05 x V _{IO}		V
SCL, SDA Logic Input Current		V _{IO} = 3.6V	-10		+10	μA
SCL, SDA Input capacitance				10		pF
SDA Output Low Voltage		Sinking 20mA			0.4	V
Output Low Voltage INTB		I _{SINK} = 1mA			0.4	V
I ² C-COMPATIBLE INTERFACE TIME	IING FOR STA	NDARD, FAST, AND FAST-MODE PL	US (Note 2)			
Clock Frequency	f _{SCL}				1000	kHz
Hold Time (Repeated) START Condition	t _{HD;STA}		0.26			μs
CLK Low Period	t _{LOW}		0.5			μs
CLK High Period	tHIGH		0.26			μs
Setup Time Repeated START Condition	tsu;sta		0.26			μs
DATA Hold Time	t _{HD:DAT}		0			μs
DATA Valid Time	t _{VD:DAT}				0.45	μs
DATA Valid Acknowledge Time	t _{VD:ACK}				0.45	μs
DATA Setup Time	t _{SU;DAT}		50			ns
Setup Time for STOP Condition	t _{SU;STO}		0.26			μs
Bus Free Time Between STOP and START	t _{BUF}		0.5			μs
Pulse Width of Spikes that Must Be Suppressed by the Input Filter		(Note 3)		50		ns

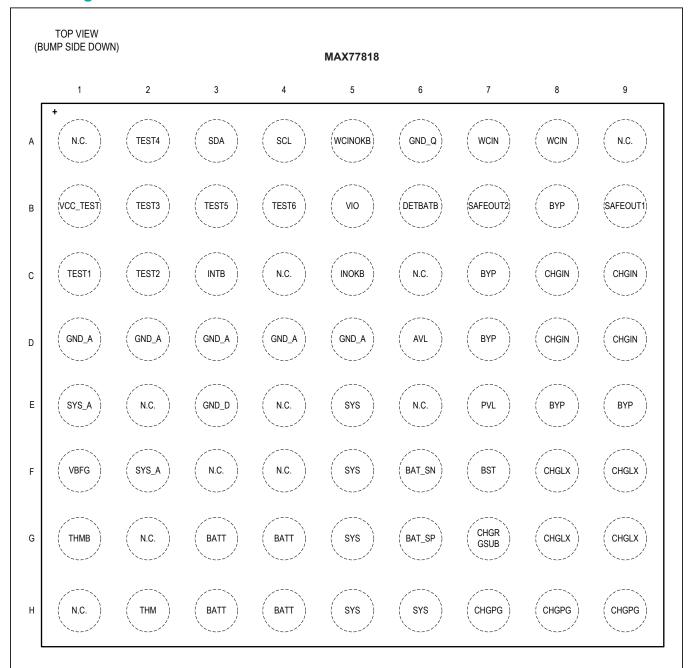
I²C Electrical Characteristics (continued)

 $(V_{SYS} = +3.7V, V_{CHGIN} = 0V, V_{IO} = 1.8V, T_A = -40^{\circ}C$ to $+85^{\circ}C$, unless otherwise noted. Limits are 100% production tested at $T_A = +25^{\circ}C$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Typical values are not guaranteed.)

DADAMETED	SYMBOL	CONDITIONS	С	C _B = 100pF		
PARAMETER			MIN	TYP	MAX	UNITS
I ² C-COMPATIBLE INTERFACE TIM	ING FOR HS MC	DDE (Note 2)				
Clock Frequency	f _{SCL}				3.4	MHz
Setup Time Repeated START Condition	t _{SU;STA}		160			ns
Hold Time (Repeated) START Condition	t _{HD;STA}		160			ns
CLK Low Period	t _{LOW}		160			ns
CLK High Period	t _{HIGH}		60			ns
DATA Setup time	t _{SU;DAT}		10			ns
DATA Hold Time	t _{HD:DAT}		0			ns
SCL Rise Time	t _{RCL}	T _A = +25°C	10		40	ns
Rise Time of SCL Signal After a Repeated START condition and After an Acknowledge Bit	^t RCL1	T _A = +25°C	10		80	ns
SCL Fall Time	t _{FCL}	T _A = +25°C	10		40	ns
SDA Rise Time	t _{RDA}	T _A = +25°C	10		80	ns
SDA Fall Time	t _{FDA}	T _A = +25°C			80	ns
Setup Time for STOP Condition	tsu;sто		160			ns
Pulse Width of Spikes that Must be Suppressed by the Input Filter				10		ns

- Note 2: Design guidance only, not tested during final test.
- Note 3: Input filters on the SDA and SCL inputs suppress noise spikes of less than 50ns.
- Note 4: The CHGIN input must be less than V_{OVLO} and greater than both $V_{CHGIN\ UVLO}$ and $V_{CHGIN2SYS}$ for the charger to turn-on.
- Note 5: The input voltage regulation loop decreases the input current to regulate the input voltage at V_{CHGIN_REG}. If the input current is decreased to I_{CHGIN_REG} of and the input voltage is below V_{CHGIN_REG}, then the charger input is turned off.
- **Note 6:** Production tested to $\frac{1}{4}$ of the threshold with LPM bit = 1 ($\frac{1}{4}$ FET configuration).
- Note 7: Production tested in charger DC-DC low-power mode.
- Note 8: Not production tested.
- Note 9: Not production tested.
- Note 10: The total chip supply current includes the charger supply current in addition to the supply current for the fuel gauge.
- **Note 11:** Symmetrical error is the sum of odd order errors in the measured values at two inputs symmetrical around zero; for example, ISERR 0.3A = (Error 0.3A Error -0.3A)/2/0.3A x 100.
- **Note 12:** Total current measurement error is the sum of the symmetrical and asymmetrical errors. Fuel gauge accuracy is sensitive to asymmetrical error but insensitive to symmetrical error.
- Note 13: Current and ratiometric measurement errors are production tested at V_{SYS} = 3.7V and guaranteed by design at V_{SYS} = 2.8V and 4.5V.
- **Note 14:** Asymmetrical error is the sum of even order errors in the measured values at two inputs symmetrical around zero; for example IAERR_0.3A = (Error 0.3A + Error -0.3A)/2.
- **Note 15:** Total linear regulator mode current measurement error is simply the total error with respect to the input. This mode exists for a short duration when charging an empty battery, hence this error has limited consequence.

Pin Configuration



WLP (72 BUMPS WLP 9X8 BUMP ARRAY 0.4MM PITCH)

N/C PINS ARE FLOATING AND CAN BE CONNECTED AT BOARD-LEVEL IF NEEDED.

ALL TEST PINS (TEST1-6 AND VCCTEST) SHOULD BE GROUNDED IN THE END-USE APPLICATION.

*TOP VIEW = WAFER BACK-SIDE VIEW (BUMPS NOT VIEWABLE)

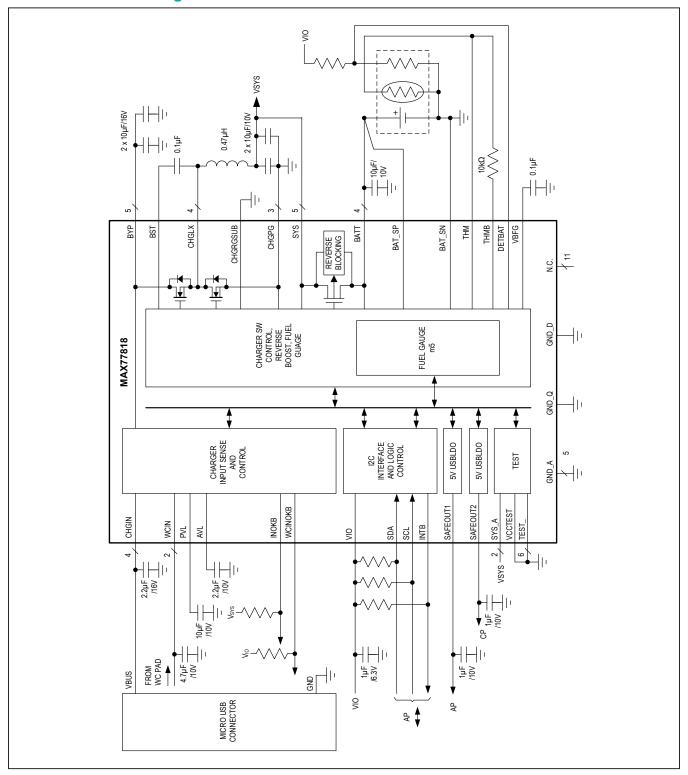
Pin Description

PIN	NAME	FUNCTION		
A1, A9, C4, C6, E2, E4, E6, F3, F4, G2, H1	N.C.	No Connection		
A2	TEST4	Test I/O Pin. Ground this pin in the application.		
A3	SDA	I ² C Serial Data. Add an external 2.2k Ω pullup resistor to V _{IO} .		
A4	SCL	I^2C Serial Clock. Add an external 2.2k $Ω$ pullup resistor to V_{IO} .		
A5	WCINOKB	Open-Drain, Active-Low WCIN Input Valid Indication Output. Pulls low when a valid WCIN is available and SYS is in the valid range. Pulls high with an off-chip pullup resistor to V_{IO} when WCIN is invalid.		
A6	GND_Q	Quiet Ground. Short to GND_A and GND_D.		
A7, A8	WCIN	6 V _{DC} Protected Input for USB or Wireless Charging. Bypass with 4.7μF/10V ceramic capacitor to GND.		
B1	V _{CCTEST}	Test Mux Supply. Ground this pin in the application.		
B2	TEST3	Test I/O. Ground this pin in the application.		
В3	TEST5	Test I/O. Ground this pin in the application.		
B4	TEST6	Test I/O. Ground this pin in the application.		
B5	V _{IO}	Digital I/O Supply Input for I ² C Interface.		
B6	DETBATB	Active-Low Battery Detection Input. Connected to V_{IO} through an off-chip pullup resistor or left unconnected disables charging. Pulled low below 80% of V_{IO} enables charging.		
B7	SAFEOUT2	High-Voltage Input LDO SAFEOUT2 Output. Default off. Bypass with 1μF/10V ceramic capacitor to GND.		
B8, C7, D7, E8, E9	ВҮР	Switching Charger Buck Input and Reverse Boost Output. Bypass with 2x10µF/25V ceramic capacitors to CHGPG.		
В9	SAFEOUT1	High-Voltage Input LDO SAFEOUT1 Output. Default 4.9V, on when CHGIN is valid. Bypass with 1μF/10V ceramic capacitor to GND.		
C1	TEST1	Test I/O. Ground this pin in the application.		
C2	TEST2	Test I/O. Ground this pin in the application.		
C3	INTB	Interrupt Output. Active-low, open-drain output. Add a 200kΩ pullup resistor to V _{IO} .		
C5	INOKB	Open-Drain, Active-Low CHGIN Input Valid Indication Output. Pulls low when a valid CHGIN is available and SYS is in the valid range. Pulls high with an off-chip pullup resistor to V _{IO} when CHGIN is invalid.		
C8, C9, D8, D9	CHGIN	High Voltage Input for USB or Wireless Charging. Bypass with 2.2µF/25V ceramic capacitor to GND.		
D1-D5	GND_A	Analog Ground. Short to GND_D and GND_Q.		

Pin Description (continued)

PIN	NAME	FUNCTION	
D6	AVL	Output of an integrated 5V LDO powering on-chip, low noise circuits. Powering external loads other than pullup resistors is not recommended. Bypass with 2.2µF/10V ceramic capacitor to GND.	
E1, F2	SYS_A	Analog SYS Input	
E3	GND_D	Digital Ground. Short to GND_A and GND_Q.	
E5, F5, G5, H5, H6	SYS	System Power Connection, Output of Charger Buck. Bypass with 2x10µF/10V ceramic capacitors to CHGPG.	
E7	PVL	Output of an Integrated 5V LDO Powering On-Chip, High Current Circuits. Powering external loads is not recommended. Bypass with 10µF/10V ceramic capacitor to GND.	
F1	V_{BFG}	Output of an Integrated 1.8V Power Supply for Fuel Gauge. Powering external load is not recommended. Bypass with 0.1µF/6.3V ceramic capacitor to GND.	
F6	BAT_SN	Battery Negative Differential Sense Connection. Connect to the negative or ground terminal close to the battery.	
F7	BST	High-Side FET Driver Supply. Bypass BST to LX with a 0.1µF ceramic capacitor.	
F8, F9, G8, G9	CHGLX	Charger Switching Node. Connect the inductor between CHGLX and SYS.	
G1	THMB	Pullup Voltage for THM Pin Pullup Resistor. Can be switched to save power.	
G3, G4, H3, H4	BATT	Battery Power Connection. Connect to the positive terminal of a single-cell (or parallel cell) Li Ion battery. Bypass BATT to CHGPG ground plane with a 10µF/10V ceramic capacitor.	
G6	BAT_SP	Battery Positive Differential Sense Connection. Connect to the positive terminal close to the battery.	
G7	CHGRGSUB	Substrate Charger Ground Connection. Connect with GND_A.	
H2	ТНМ	Thermistor Connection. Determines battery temperature using ratiometric measurement.	
H7–H9	CHGPG	Charger Power Ground Connection.	

Functional Block Diagram



Detailed Description

Switching Charger

The MAX77818 includes a full-featured switch-mode charger for a one-cell lithium ion (Li+) or lithium polymer (Li-poly) battery. Figure 2 shows the detailed functional diagram. One high voltage input (CHGIN) and one low voltage input (WCIN) can be enabled or disabled independently with I²C. When enabled, the input powers the system and charges the battery. Both inputs have programmable current limit and under-/over-voltage protections. The reverse leakage blocking feature prevents back powering from the battery or between the two inputs.

The synchronous switch-mode DC-DC converter utilizes a high 4.0MHz switching frequency which is ideal for portable devices because it allows the use of small components while eliminating excessive heat generation. The DC-DC has both a buck and a boost mode of operation. When a valid input is available (CHGIN or WCIN), the DC-DC operates as a buck, delivering energy to the system and charge the battery. When no valid inputs are available and boost mode or OTG mode is enabled, the DC-DC operates in reverse mode, and outputs a boosted voltage at BYP powered by battery. CHGIN can be used to power USB On-the-Go (OTG) devices when OTG mode is enabled.

Maxim's Smart Power Selector™ architecture makes the best use of the limited adapter power and the battery's power at all times to supply up to 3.0A continuous (4A peak) from the buck to the system. Additionally, supplement mode provides additional current from the battery to the system up to 4.5A_{RMS}. Adapter power that is not used for the system goes to charge the battery. All power switches for charging and switching the system load between battery and adapter power are integrated on chip. No external MOSFETs are required.

Maxim's proprietary process technology allows for low-RDSON devices in a small solution size. The total dropout resistance from CHGIN to BATT is 0.0599Ω excluding inductor DCR, allowing high-efficiency and low heat operation.

A multitude of safety features ensures reliable charging. Features include charge timers, watchdog, junction thermal regulation, over-/under-voltage protection, and short-circuit protection.

ModelGauge m5 Fuel Gauge

The MAX77818 is the latest high-performance battery charger that integrates the Maxim proprietary ModelGauge m5 algorithm in the same chip. Classical coulomb-counter-based fuel gauges have excellent linearity and shortterm performance. However, they suffer from drift due to the accumulation of the offset error in the current-sense measurement. Although the offset error is often very small, it cannot be eliminated, causing the reported capacity error to increase over time, which requires periodic corrections. The ModelGauge m5 algorithm does not suffer from abrupt corrections that normally occur in coulomb-counter algorithms, since tiny continual corrections are distributed over time. The ModelGauge m5 adopts a mixing algorithm that combines the VFG capacity with the coulomb counter and weighs each result so that both are used optimally to determine the battery state. In this way, the VFG capacity result is used to continuously make small adjustments to the battery state, canceling the coulomb-counter drift.

The MAX77818 includes improved age adaptation, improved SOC accuracy to empty, and increased temperature measurement. The device provides two methods for reporting the age of the battery: reduction in capacity and cycle odometer. Moreover, the device provides precision measurements of current, voltage, and temperature. Details are discussed in the <u>ModelGauge m5 Details</u> section.

Smart Power Selector is a trademark of Maxim Integrated Products, Inc.

Charger Details

Smart Power Selector (SPS)

The Smart Power Selector architecture includes a network of internal switches and control loops that distributes energy among CHGIN, WCIN, BYP, SYS, and BATT.

 $\underline{\text{Figure 2}}$ shows the Smart Power Selector switches: QCHGIN, QWCIN, QHS, QLS, and QBAT.

Switch and Control Loop Descriptions:

CHGIN Input Switch: QCHGIN is used for input current sensing. It turns completely on when a valid CHGIN is available and does not provide forward blocking. SPS control loops regulate CHGIN input current and voltage.

DC-DC Switches: Q_{HS} and Q_{LS} are the DC-DC switches that can operate as a buck (step-down) or a boost (step-up). When operating as a buck, energy is delivered from BYP to SYS. When operating as a boost, energy is delivered from SYS to BYP. SPS control loops monitor the DC-DC switch current, the SYS voltage, and the BYP voltage.

Battery-to-System Switch: Q_{BAT} controls the battery charging and discharging. Additionally, Q_{BAT} allows the battery to be isolated from the system (SYS). SPS control loops regulate battery current and voltage.

Control Bits

MODE[3:0] configures the Smart Power Selector.

MINVSYS[1:0] sets the minimum system voltage.

VBYPSET[6:0] sets the BYP regulation voltage target.

B2SOVRC[2:0] sets the battery overcurrent protection threshold.

Energy Distribution Priority:

With a valid external power source (buck operation):

- The external power source is the primary source of energy.
- The battery is the secondary source of energy.
- Energy delivery to SYS is the highest priority.
- Any energy that is not required by SYS is available to the battery charger.

With no valid external power source:

- The battery is the source of energy.
- Energy delivery to SYS is the highest priority.
- Any energy not required by SYS is available for BYP or CHGIN (when boost or OTG is enabled).

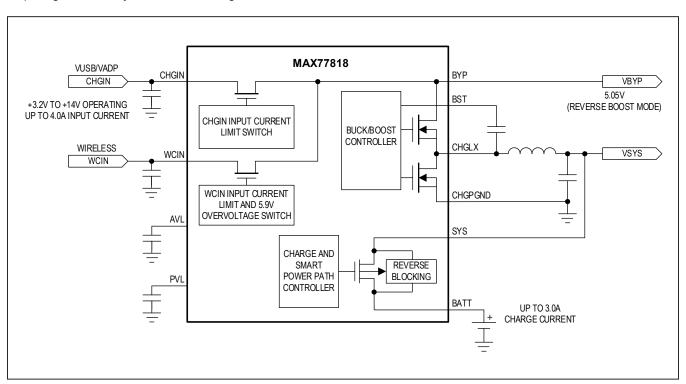


Figure 1. Simplified Charger Functional Diagram

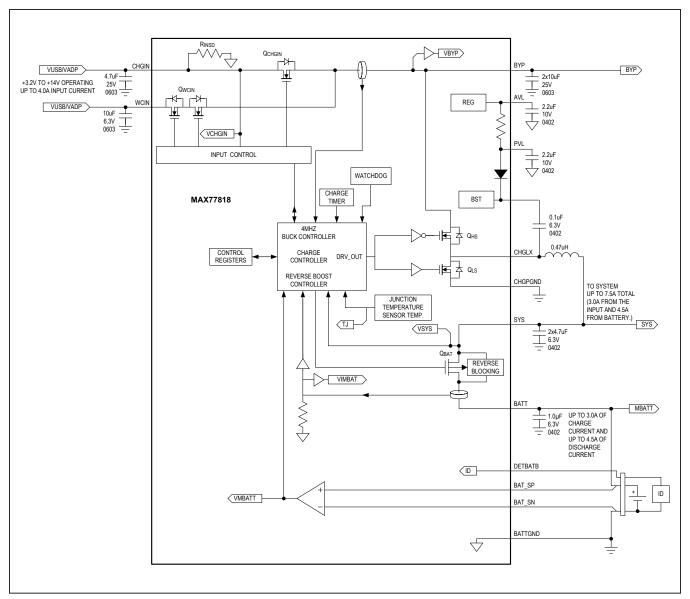


Figure 2. Battery Charger Detailed Functional Diagram

BYP Regulation Voltage

When the DC-DC is enabled in boost only mode (MODE[3:0] = 0x08), the voltage from BYP to ground (V_{BYP}) is regulated to VBYPSET[6:0].

When the DC-DC is enabled in one of its USB OTG modes (MODE[3:0] = 0x0A or 0x0E or 0x0F), V_{BYP} is set for 5.1V ($V_{BYP.OTG}$).

When the DC-DC is off or in one of its buck modes (MODE[3:0] = 0x00 or 0x04 or 0x05) and there is a valid power source at CHGIN, $V_{BYP} = V_{CHGIN} - I_{CHGIN} \times R_{QCHGIN}$. When the DC-DC is off and there is no valid power source at CHGIN, BYP is connected to SYS with an internal 200Ω resistor. This 200Ω resistor keeps BYP biased at V_{SYS} and allows for the system to draw very light loads from BYP. If the system loading on BYP is more than 1.0mA then the DC-DC should be operated in boost mode. Note that the inductor and the high-side switch's body diode are in parallel with the 200Ω from SYS to BYP.

SYS Regulation Voltage

When the DC-DC is enabled as a buck and the charger is disabled (MODE[3:0] = 0x04), V_{SYS} is regulated to V_{BATREG} (CHG_CV_PRM) and Q_{BAT} is off.

When the DC-DC is enabled as a buck and the charger is enabled but in a non-charging state such as done, watchdog suspend or timer fault (MODE[3:0] = 0x05 and not charging), V_{SYS} is regulated to V_{BATREG} (CHG_CV_PRM) and Q_{BAT} is off.

When the DC-DC is enabled as a buck and charging in prequalification, fast-charge, or top-off modes (MODE[3:0] = 0x05 and charging), V_{SYS} is regulated to V_{SYSMIN} when the $V_{BATT} < V_{SYSMIN}$; in this mode the Q_{BAT} switch acts like a linear regulator and dissipates power $[P = (V_{SYSMIN} - V_{BATT}) \times I_{BATT}]$. When $V_{BATT} \times V_{SYSMIN}$, then $V_{SYS} = V_{BATT} - I_{BATT} \times R_{BAT2SYS}$; in this mode the Q_{BAT} switch is closed.

In all of the above modes, if the combined SYS and BYP loading exceeds the input current limit, then the battery provides supplemental current to the system, and V_{SYS} regulates to V_{BATT} - V_{BSREG} . If the fuel gauge requests battery information (voltage and current) during supplement mode, then the Q_{BAT} switch is closed (V_{SYS}

= V_{BATT} - I_{BATT} x $R_{BAT2SYS}$) during the fuel gauge sampling period. If the fuel gauge requests continuous samples from the battery during supplement mode, then the Q_{BAT} switch eventually opens when I_{BATT} decreases below 40mA.

When the DC-DC is enabled in boost or OTG modes (MODE[3:0] = 0x08, 0x0A, 0x0C, 0x0D, 0x0E, or 0x0F), then the QBAT switch is closed and $V_{SYS} = V_{BATT} - I_{BATT} \times R_{BAT2SYS}$.

Battery Detect Input Pin (DETBATB)

Utilize the DETBATB pin to disable charging until the host processor configures the charger, or connect the DETBATB pin to the ID pin of the battery pack and allow charging to automatically start when a valid charging source is available (Figure 3).

Input Validation

As shown in <u>Figure 4</u>, the charger input is compared with several voltage thresholds to determine if it is valid. A charger input must meet the following three criteria to be valid:

- 1) CHGIN must be above V_{CHGIN_UVLO} to be valid.
- 2) CHGIN must be below its overvoltage-lockout threshold (V_{CHGIN-OVLO}).
- 3) CHGIN must be above the system voltage by VCHGIN2SYS.

CHGIN input generates a CHGIN_I interrupt when its status changes. The input status can be read from CHGIN_OK and CHGIN_DTLS register bits. Interrupts can be masked with CHGIN M.

The WCIN input validation is similar.

Input Current Limit

Set CHGIN and WCIN input current regulation thresholds with CHGIN_ILIM[6:0] and WCIN_ILIM[5:0] register bits. The default settings of the CHGIN_ILIM[6:0], WCIN_ILIM[5:0] and MODE[3:0] register bits are such that when a charge source is applied to CHGIN or WCIN, the MAX77818 turns its DC-DC converter on in BUCK mode, limits $V_{\mbox{\footnotesize{SYS}}}$ to $V_{\mbox{\footnotesize{BATREG}}}$, and limits the charge source current to 500mA. All control bits are reset on global shutdown.

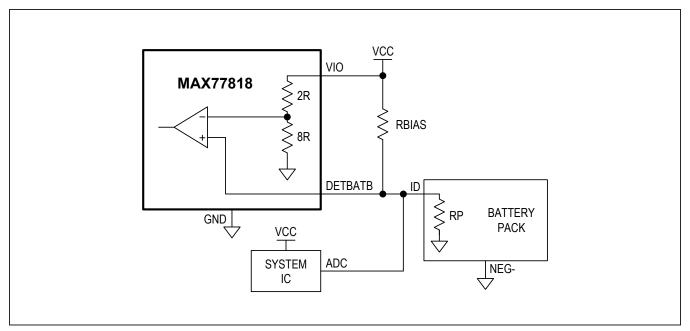


Figure 3. DETBATB Internal Circuitry and System Diagram

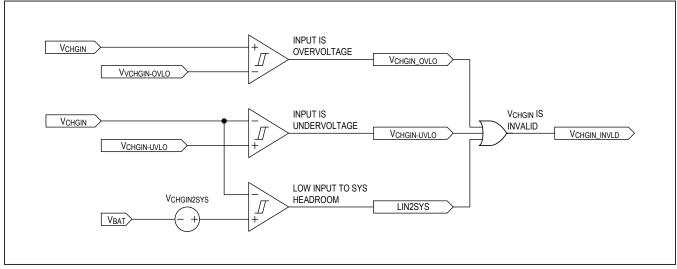


Figure 4. Charger Input Validation

Input-Voltage Regulation (V_{CHGIN REG})

The input-voltage regulation loop allows the charger to extract maximum input power while avoiding input source overload when it is attached to a poor quality power source (CHGIN) or wireless charger (WCIN). The loop improves performance with relatively high-resistance charge sources that exist when long cables are used or devices are charged with noncompliant USB hub configurations. Additionally, this input-voltage regulation loop improves performance with current limited adapters. If the MAX77818's input current limit is programmed above the current limit threshold of given adapter, the input voltage loop allows the MAX77818 to regulate at the current limit of the adapter. Finally, the input-voltage regulation loop allows the MAX77818 to perform well with adapters that have poor transient load response times.

The input-voltage regulation loop automatically reduces the input current in order to regulate the input voltage at VCHGIN_REG or VWCIN_REG. If the input current is reduced to ICHGIN_REG_OFF (50mA typ) and the input voltage is below VCHGIN_REG, then QCHGIN is turned off. VCHGIN_REG and VWCIN_REG are programmable with VCHGIN_REG[1:0] and VWCIN_REG[1:0] register bits.

After input-voltage regulation is activated, a BYP_I interrupt is generated, BYP_OK is cleared and BYP_DTLS[3:0] = 0b1xxx. To optimize input power when working with a current limited charge source, monitor BYP_DTLS[3:0] while decreasing the input current limit with CHGIN_ILIM[6:0]. When the input current limit is set below the current limit of the adapter, the input voltage rises. Although the input current limit is lowered, more power can be extracted from the input source when the input voltage is allowed to rise.

Example 1: Optimum use of the input-voltage regulation loop along with a current limited adapter.

Sequence of Events

 V_{BATT} = 3.2V, the system is operating normally.

A 5.0V 1.2A current limited dedicated USB charger is applied to CHGIN.

The system detects that the charge source is a dedicated USB charger and enables the battery charger (MODE = 0x05) and programs an input current limit to 1.8A (CHGIN_ILIM = 0x36 = 1.8A).

The input current starts to ramp up from 100mA to 1.8A, but at the input current limit of the adapter (1.2A), the adapter voltage collapses. The MAX77818's input-voltage regulation loop prevents the adapter voltage from falling below 4.3V (V_{CHGIN_REG} = 4.3V). A BYP_I interrupt is generated.

With the input-voltage regulation loop active, the adapter provides 1.2A at 4.3V, which is a total of 5.04W being delivered to the system.

The system software detects that the input-voltage regulation loop is active and it begins to ramp down the programmed input current limit. When the current limit ramps down to 1.167A (CHGIN_ILIM[6:0]), the adapter is no longer in current limit and the adapter voltage increases from 4.3V to 5.0V.

When the adapter operates just below its current limit, it provides 1.167A at 5.0V which is a total of 5.84W to the system. This is 800mW more than when the adapter was in current limit.

Input Self-Discharge for Reliable Charger Input Interrupt

To ensure that a rapid removal and reinsertion of a charge source always results in a charger input interrupt, the charger input presents loading $R_{\mbox{\footnotesize{INSD}}}$ to the input capacitor to ensure that when the charge source is removed, the input voltage decays below the UVLO threshold in a reasonable time

System Self-Discharge with No Power

To ensure a timely, complete, repeatable, and reliable reset behavior when the system has no power, the MAX77818 actively discharges the BATT and SYS nodes when the battery is removed and $V_{\mbox{\footnotesize SYS}}$ is less than $V_{\mbox{\footnotesize SYSUVLO}}.$ As shown in $\underline{\mbox{\footnotesize Figure 5}},$ the BATT and SYS discharge resistors are both $600\Omega.$

Charge States

The MAX77818 utilizes several charging states to safely and quickly charge batteries as shown in Figure 6 and Figure 7. Figure 6 shows an exaggerated view of a Li+/Li-Poly battery progressing through the following charge states when there is no system load and the die and battery are close to room temperature: daed battery/low battery prequalification → fast-charge → top-off → done.

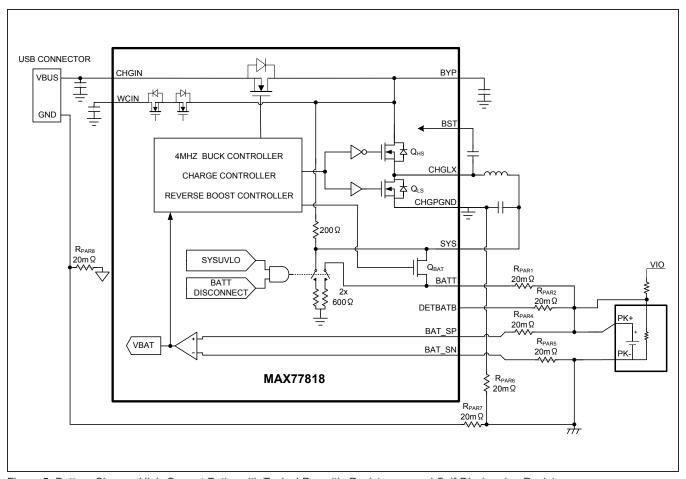


Figure 5. Battery Charger High-Current Paths with Typical Parasitic Resistances and Self-Discharging Resistors

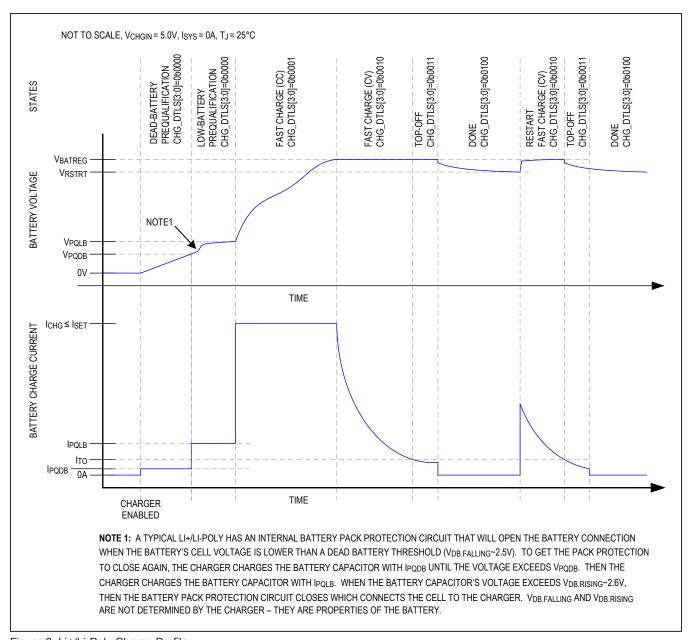


Figure 6. Li+/Li-Poly Charge Profile

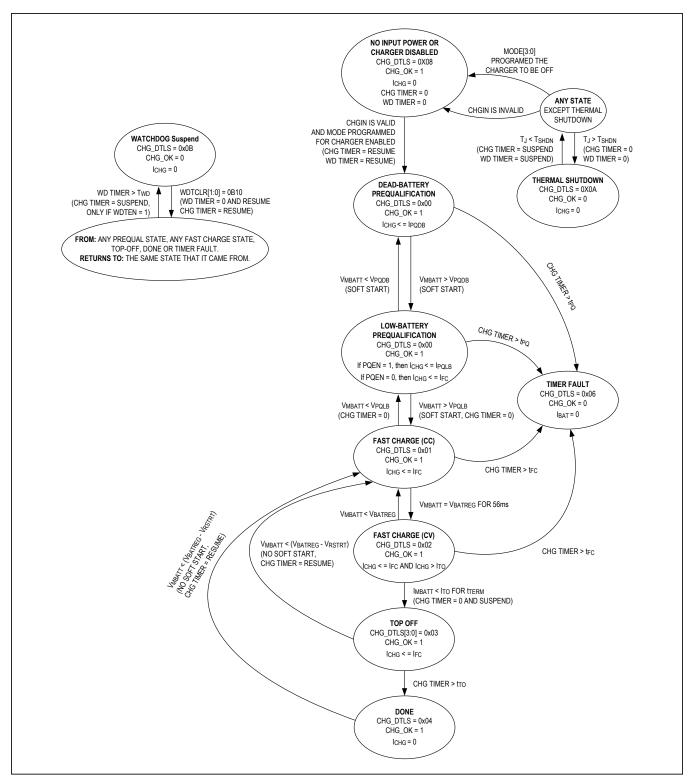


Figure 7. Charger State Diagram

No Input Power or Charger Disabled State

From any state shown in <u>Figure 7</u> except thermal shutdown, the no input power or charger disabled state is entered whenever the charger is programmed to be off or the charger input CHGIN is invalid. After being in this state for t_{SCIDG}, a CHG_I interrupt is generated, CHG_OK is set, and CHG_DTLS[3:0] is set to 0x08.

While in the no input power or charger disabled state, the charger current is 0mA, the watchdog and charge timers are forced to 0, and the power to the system is provided by either the battery or the adapter. When both battery and adapter power are available, the adapter provides primary power to the system and the battery contributes supplemental energy to the system if necessary.

To exit the no input power or charger disabled state, the charger input must be valid and the charger must be enabled.

Dead Battery Prequalificiation State

As shown in Figure 7, the dead battery prequalification state occurs when the main battery voltage is less than V_{PQDB} . After being in this state for t_{SCIDG} , a CHG_I interrupt is generated, CHG_OK is set, and CHG_DTLS[3:0] is set to 0x00. In the dead battery prequalification state, charge current into the battery is I_{PQDB} .

The following events cause the state machine to exit this state:

The main battery voltage rises above V_{PQDB} and the charger enters the next state in the charging cycle: low battery pregualification.

If the battery charger remains in this state for longer than t_{PQ} , the charger state machine transitions to the timer fault state.

If the watchdog timer is not serviced, the charger state machine transitions to the watchdog suspend state.

Note that the dead battery prequalification state works with battery voltages down to 0V. The low 0V operation typically allows this battery charger to recover batteries that have an open internal pack protector. Typically, a packs internal protection circuit opens if the battery has seen an over current, undervoltage, or overvoltage. When a battery with an open internal pack protector is used with this charger, the low battery prequalification mode current flows into the 0V battery. This current raises the pack's terminal voltage to the point where the internal pack protection switch closes.

Note that a normal battery typically stays in the low battery prequalification state for several minutes or less.

Therefore, a battery that stays in low battery prequalification for longer than t_{PQ} may be experiencing a problem.

Fast-Charge Constant Current State

As shown in Figure 7, the fast-charge constant current (CC) state occurs when the main battery voltage is greater than the low battery prequalification threshold and less than the battery regulation threshold ($V_{PQLB} < V_{BATT} < V_{BATREG}$). After being in the fast-charge CC state for t_{SCIDG}, a CHG_I interrupt is generated, CHG_OK is set and CHG_DTLS[3:0] = 0x01.

In the fast-charge CC state, the current into the battery is less than or equal to I_{FC} . Charge current can be less than I_{FC} for any of the following reasons:

- The charger input is in input current limit.
- The charger input voltage is low.
- The charger is in thermal foldback.

The system load is consuming adapter current. Note that the system load always gets priority over the battery charge current.

The following events cause the state machine to exit this state:

When the main battery voltage rises above V_{BATREG}, the charger enters the next state in the charging cycle: fast charge (CV).

If the battery charger remains in this state for longer than t_{FC} , the charger state machine transitions to the timer fault state.

If the watchdog timer is not serviced, the charger state machine transitions to the watchdog suspend state.

The battery charger dissipates the most power in the fast-charge constant current state. This power dissipation causes the internal die temperature to rise. If the die temperature exceeds T_{REG} , I_{FC} is reduced.

Fast-Charge Constant Voltage State

As shown in Figure 7, the fast-charge constant voltage (CV) state occurs when the battery voltage rises to V_{BATREG} from the fast-charge CC state. After being in the fast-charge CV state for t_{SCIDG} , a CHG_I interrupt is generated, CHG_OK is set, and CHG_DTLS[3:0] = 0x02.

In the fast-charge CV state, the battery charger maintains V_{BATREG} across the battery and the charge current is less than or equal to I_{FC} . As shown in Figure 6, charger current decreases exponentially in this state as the battery becomes fully charged.

The Smart Power Selector control circuitry may reduce the charge current lower than the battery may otherwise consume for any of the following reasons:

- The charger input is in input current limit.
- The charger input voltage is low.
- The charger is in thermal foldback.

The system load is consuming adapter current. Note that the system load always gets priority over the battery charge current.

The following events cause the state machine to exit this state:

When the charger current is below I_{TO} for t_{TERM} , the charger enters the next state in the charging cycle: TOP OFF.

If the battery charger remains in this state for longer than t_{FC} , the charger state machine transitions to the timer fault state.

If the watchdog timer is not serviced, the charger state machine transitions to the watchdog suspend state.

Top-Off State

As shown in Figure 7, the top-off state can only be entered from the fast-charge CV state when the charger current decreases below I_{TO} for t_{TERM} . After being in the top-off state for t_{SCIDG} , a CHG_I interrupt is generated, CHG_OK is set, and CHG_DTLS[3:0] = 0x03. In the top-off state, the battery charger tries to maintain V_{BATREG} across the battery and typically the charge current is less than or equal to I_{TO} .

The Smart Power Selector control circuitry can reduce the charge current lower than the battery is able to. Otherwise, consume for any of the following reasons:

- The charger input is in input current limit.
- The charger input voltage is low.
- The charger is in thermal foldback.

The system load is consuming adapter current. Note that the system load always gets priority over the battery charge current.

Following events causes the state machine to exit this state:

After being in this state for the top-off time (t_{TO}), the charger enters the next state in the charging cycle: DONE.

If $V_{BATT} < V_{BATREG} - V_{RSTRT}$, the charger goes back to the FAST CHARGE (CC) state.

If the watchdog timer is not serviced, the charger state machine transitions to the watchdog suspend state.

Done State

As shown in Figure 7, the battery charger enters its done state after the charger has been in the top-off state for t_{TO} . After being in this state for t_{SCIDG} , a CHG_I interrupt is generated, CHG_OK is cleared, and CHG_DTLS[3:0] = 0x04.

The following events cause the state machine to exit this state:

If $V_{BATT} < V_{BATREG} - V_{RSTRT}$, the charger goes back to the fast charge (CC) state.

If the watchdog timer is not serviced, the charger state machine transitions to the watchdog suspend state.

In the done state, the charge current into the battery (I_{CHG}) is 0A. In the done state, the charger presents a very low load (I_{MBDN}) to the battery. If the system load presented to the battery is low (<< 100µA), then a typical system can remain in the done state for many days. If left in the done state long enough, the battery voltage decays below the restart threshold (V_{RSTRT}) and the charger state machine transitions back into the fast-charge CV state. There is no soft-start (di/dt limiting) during the done to fast-charge state transition.

Timer-Fault State

The battery charger provides both a charge timer and a watchdog timer to ensure safe charging. As shown in Figure 7, the charge timer prevents the battery from charging indefinitely. The time that the charger is allowed to remain in each of its prequalification states is $t_{PQ}.$ The time that the charger is allowed to remain in the fast-charge CC and CV states is $t_{FC},$ which is programmable with FCHGTIME. Finally, the time that the charger is in the top-off state is $t_{TO},$ which is programmable with TO_TIME. Upon entering the timer-fault state, a CHG_I interrupt is generated without a delay, CHG_OK is cleared, and CHG_DTLS = 0x06.

In the timer-fault state the charger is off. The charger can exit the timer-fault state by programming the charger to be off and then programming it to be on again through the MODE bits. Alternatively, the charger input can be removed and reinserted to exit the timer-fault state. See the any state bubble in the upper right of Figure 7.

Watchdog Timer

The battery charger provides both a charge timer and a watchdog timer to ensure safe charging. As shown in Figure 7, the watchdog timer protects the battery from charging indefinitely in the event that the host hangs or otherwise cannot communicate correctly. The watchdog timer is disabled by default with WDTEN = 0. To use the watchdog timer feature, enable the feature by setting WDTEN. While enabled, the system controller must reset the watchdog timer within the timer period (t_{WD}) for the charger to operate normally. Reset the watchdog timer by programming WDTCLR = 0x01.

If the watchdog timer expires while the charger is in dead-battery prequalification, low-battery prequalification, fast-charge CC or CV, top-off, done, or timer fault, the charging stops, a CHG_I interrupt is generated without a delay, CHG_OK is cleared, and CHG_DTLS[3:0] indicates that the charger is off because the watchdog timer expired. Once the watchdog timer has expired, the charger can be restarted by programming WDTCLR = 0x01. The SYS node can be supported by the battery and/or the adapter through the DC-DC buck while the watchdog timer is expired.

Thermal-Shutdown State

As shown in Figure 7, the thermal-shutdown state occurs when the battery charger is in any state and the junction temperature (T_J) exceeds the device's thermal-shutdown threshold (T_{SHDN}). When T_J is close to T_{SHDN} , the charger has folded back the input current limit to 0A so the charger and inputs are effectively off. Upon entering this state, CHG_I interrupt is generated without a delay, CHG_OK is cleared, and CHG_DTLS[3:0] = 0x0A.

In the thermal-shutdown state the charger is off and timers are suspended. The charger exits the temperature suspend state and returns to the state it came from once the die temperature has cooled. The timers resume once the charger exits this state.

Battery Differential Voltage Sense

As shown in <u>Figure 2</u>, BAT_SP and BAT_SN are differential remote sense voltage inputs for the battery. To improve accuracy and decrease charging time, the battery charger voltage sense is based on the differential voltage between BAT_SP and BAT_SN.

<u>Figure 5</u> shows the high-current paths of the battery charger along with some example parasitic resistances. With the parasitics described in <u>Figure 5</u>, a charge current of 1A measuring from BATT to GND leads to a V_{BATT} that is 40mV higher than the real voltage because of R_{PAR1} and R_{PAR7} (I_{CHG} x (R_{PAR1} + R_{PQR7}) = 1A x 40mΩ = 40mV).

Since the charger thinks the battery voltage is higher than it actually is, it enters its fast-charge CV state sooner and the effective charge time may be extended by 10 minutes (based on real lab measurements). This charger with differential remote sensing does not experience this type of problem because BAT_SP and BAT_SN sense the battery voltage directly. To get the maximum benefit from these sense lines, connect them as close as possible to the battery connector.

OTG Mode

The DC-DC converter topology of the MAX77818 allows it to operate as a forward buck converter or as a reverse-boost converter. The operating modes of the DC-DC converter are controlled with MODE[3:0] register bits. When MODE[3:0] = 0x0A, 0x0E or 0x0F the OTG mode is enabled, the DC-DC converter provides power to CHGIN pin, commonly known as USB On-the-Go (OTG).

When OTG mode is enabled, the DC-DC converter operates in reverse-boost mode and regulates V_{BYP} to $V_{BYP.OTG}$ (5.1V, typ) and the switch from BYP to CHGIN is closed. The current through the BYP to CHGIN switch is limited to the value programmed by OTG_ILIM[1:0].

If the external OTG load at CHGIN exceeds $I_{CHGIN.OTG.}$ I_{LIM} , then a BYP_I interrupt is generated, BYP_OK = 0, and BYP_DTLS[3:0] = 0bxxx1. In response to an overload at CHGIN during OTG mode operation, the BYP to CHGIN switch is latched off. The BYP to CHGIN switch automatically tries to retry in ~300ms. If the overload at CHGIN persists, then the switch toggles on and off with ~30ms on and ~300ms off.

Battery Overcurrent Protection During System Power-Up

The main battery overcurrent protection during system power-up feature limits the main battery to system current to I_{SYSPU} as long as V_{SYS} is less than V_{SYSPU} . This feature limits the surge current that typically flows from the main battery to the device's low-impedance system bypass capacitors during a system power-up. System power-up is anytime that energy from the battery is supplied to SYS when $V_{SYS} < V_{SYSPU}$. This system power-up condition typically occurs when a battery is hot-inserted into an otherwise unpowered device. Similarly, the system power-up condition could occur when the DISIBS bit is driven low.

When system power-up occurs due to hot insertion into an otherwise unpowered device, a small delay of (t_{SYS-PU}) is required in order for this feature's control circuits to activate. A current spike over t_{SYSPU} can occur during this time.

Battery Overcurrent Protection Due to Fault

The MAX77818 protects itself, the battery, and the system from potential damage due to excessive battery discharge current. Excessive battery discharge current may occur in a smartphone for several reasons such as exposure to moisture, a software problem, an IC failure, a component failure, or a mechanical failure that causes a short circuit. The main battery overcurrent protection feature is enabled with B2SOVRC; disabling this feature reduces the main battery current consumption by IMBOVRC.

When the battery (BATT) to system (SYS) discharge current (I_{BATT}) exceeds the programmed overcurrent threshold for at least I_{MBOVRC} , a BAT_I interrupt is generated, BAT_OK is cleared, and BAT_DTLS reports and overcurrent condition. Typically when the system's processor detects this overcurrent interrupt, it executes a housekeeping routine that tries to mitigate the overcurrent situation. If the processor cannot correct the overcurrent, then it can disable the BATT to SYS discharge path (B2S switch) by driving DISIBS bit to a logic high.

There are different scenarios of how the MAX77818 responds to setting the DISIBS bit high depending on the available power source and the state of the charger.

The MAX77818 is only powered from BATT and DISIBS bit is set.

SYS collapses and is allowed to go to 0V.

DISIBS holds state.

To exit from this state, plug in a valid input charger, then SYS is powered up, and the system wakes up.

The MAX77818 is powered from BATT and CHGIN, and the charger buck is not switching and DISIBS bit is set.

To exit from this state, plug in a valid input charger, then SYS is powered up and the system wakes up.

The MAX77818 is powered from BATT and CHGIN and the charger buck is switching and DISIBS bit is set.

The DISIBIS bit is ignored.

Thermal Management

The MAX77818 charger uses several thermal management techniques to prevent excessive battery and die temperatures.

Thermal Monitor

The user can monitor thermistor temperature using the fuel gauge and adjust the charger voltage/current as needed.

Thermal Foldback

Thermal foldback maximizes the battery charge current while regulating the MAX77818 junction temperature. As shown in Figure 8, when the die temperature exceeds the value programmed by REGTEMP (T_{JREG}), a thermal limiting circuit reduces the battery charger's target current by 105mA/°C (ATJREG). The target charge current reduction is achieved with an analog control loop (i.e., not a digital reduction in the input current). When the thermal foldback loop changes state, a CHG I interrupt is generated and the system's microprocessor may want to read the status of the thermal regulation loop through the TREG status bit. Note that the thermal foldback loop being active is not considered to be abnormal operation and the thermal foldback loop status does not affect the CHG OK bit (only information contained within CHG DTLS affects CHG_OK).

Analog Low-Noise Power Input (AVL) and PVL

As shown in <u>Figure 2</u>, AVL is a regulated output from BYPC node. AVL is the power input for the MAX77818 charger's analog circuitry. PVL has a 12.5Ω resistor internal to the MAX77818 and a $10\mu\text{F}$ ceramic capacitor external bypass capacitor to isolate noises from AVL.

Power States

The MAX77818 transitions between power states as input/battery and load conditions dictate; see Figure 9.

The MAX77818 provides seven power states and one no power state. Under power-limited conditions, the power path feature maintains SYS and USB-OTG loads at the expense of battery charge current. In addition, the battery supplements the input power when required. As shown, transitions between power states are initiated by detection/removal of valid power sources, OTG events, and under-voltage conditions. Details of the BYP and SYS voltages are provided for each state.

No Input Power, <u>MODE = undefined</u>: No input adapter or battery is detected. The charger and system is off. Battery is disconnected and charger is off.

Battery Only, <u>MODE = 0x00</u>: Adapter and wireless charger are invalid, outside the input voltage operating range (QCHGIN = off, QWCIN = off). Battery is connected to power the SYS load (QBAT = on), and boost is ready to power OTG (Boost=standby), see Figure 10. Battery Only.

Battery Boost, MODE = 0x08: Adapter and wireless inputs are invalid, outside the input voltage operating range (QCHGIN = off, QWCIN = off). Battery is connected to power the SYS load (QBAT = on), and charger is operating in Boost mode (Boost = on). See Figure 11.

Battery Boost (OTG), <u>MODE = 0x0A</u>: Wireless input is turned off (QWCIN = off) and OTG is active (QCHGIN = on). Battery is connected to support SYS and OTG loads (QBAT = on), and charger is operating in boost mode (boost = on). See Figure 12.

No Charge Buck, <u>MODE = 0x0C</u>: Adapter or wireless charger are detected, within the input voltage operating range (QCHGIN = on or QWCIN = on). Battery is disconnected (QBAT = off), and charger is operating in buck mode powering the SYS node. See Figure 13.

Charge Buck, <u>MODE = 0x0D</u>: Adapter or wireless charger are detected, within the input voltage operating range (QCHGIN = on or QWCIN = on). Battery is connected in

charge mode (QBAT = on), and charger is operating in buck mode. See Figure 14.

No Change Buck (OTG), <u>MODE = 0x0E</u>: Wireless charger is detected within the input voltage operating range (QWCIN = on) and OTG is active (QCHGIN = of). Battery is connected in charge mode (QBAT = on), and charger is operating in buck mode. See Figure 15.

Charge Buck (OTG), <u>MODE = 0x0F</u>: Wireless charger is detected within the input voltage operating range (QWCIN = on) and OTG is active (QCHGIN = on). Battery is connected in charge mode (QBAT = on), and charger is operating in buck mode powering the SYS node. See Figure 16.

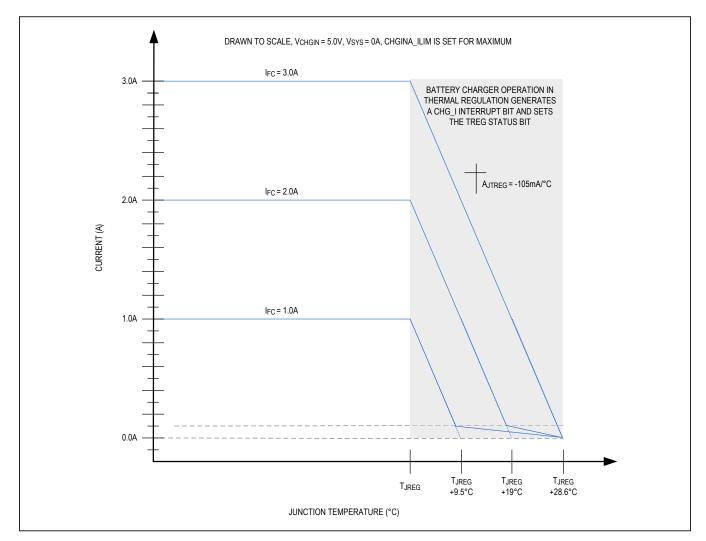


Figure 8. Charge Currents vs. Junction Temperature

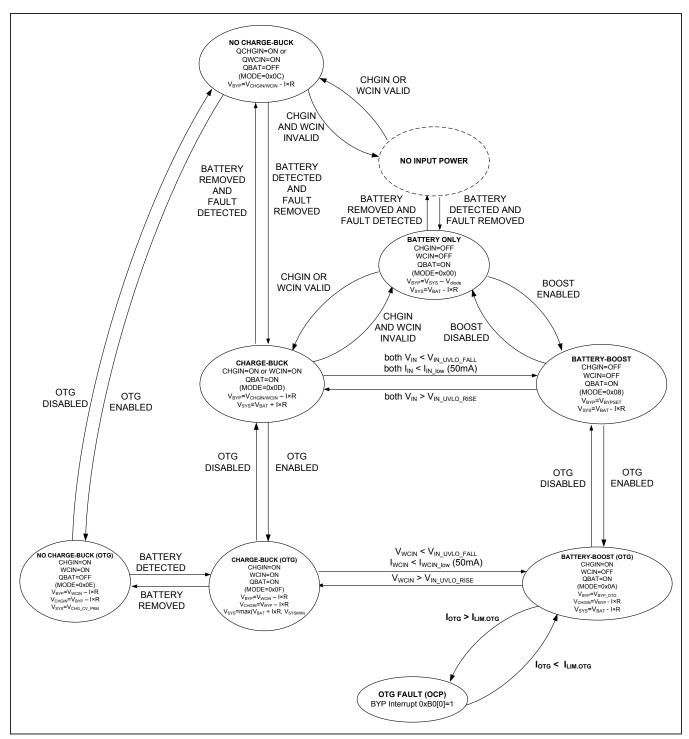


Figure 9. Power State Diagram

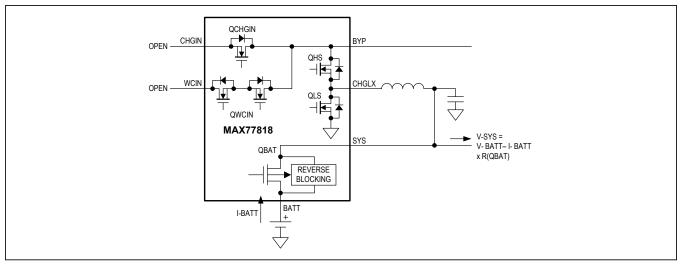


Figure 10. Battery-Only

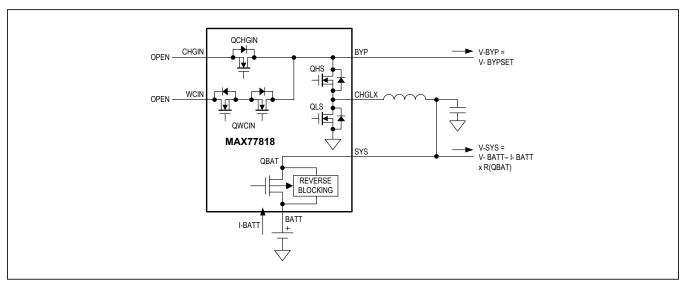


Figure 11. Battery-Boost

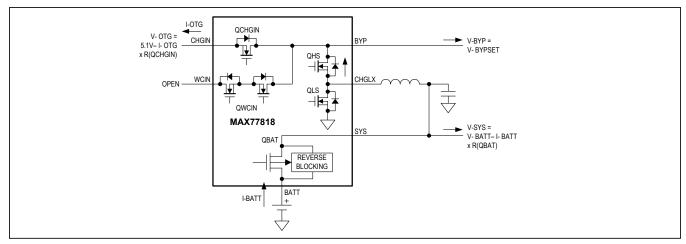


Figure 12. Battery Boost (OTG)

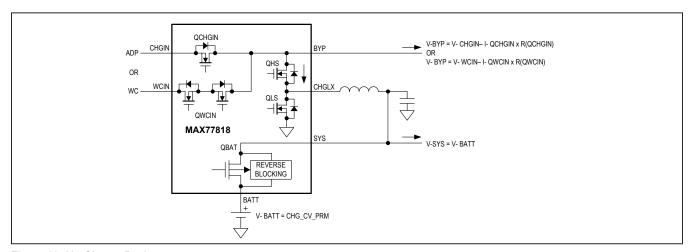


Figure 13. No Charge Buck

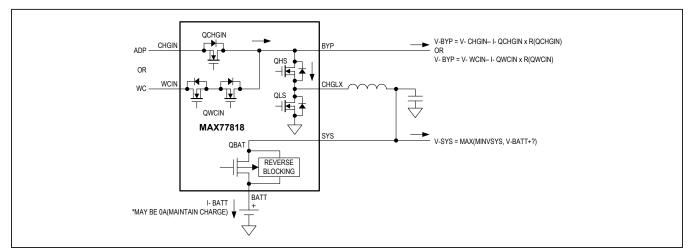


Figure 14. Charge Buck

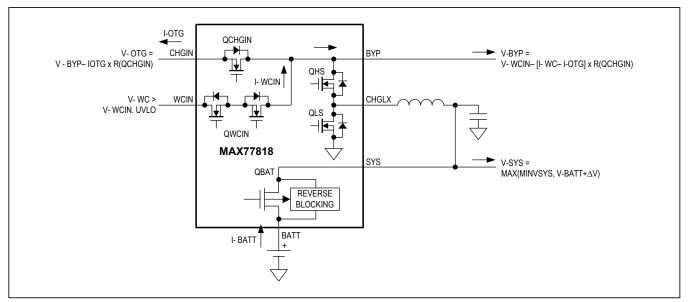


Figure 15. No Charge Buck (OTG)

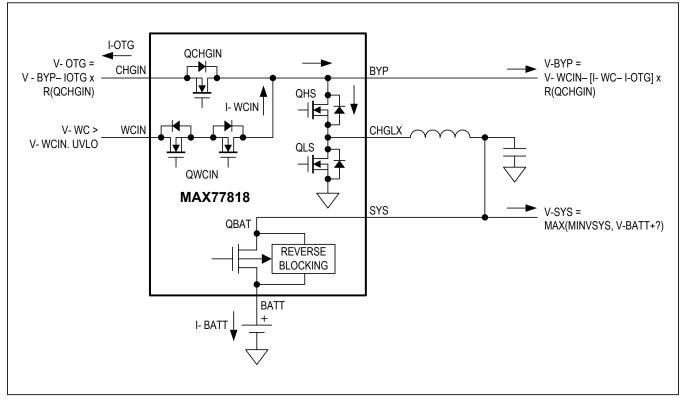


Figure 16. Charge-Buck (OTG)

ModelGauge m5 Details

The MAX77818 incorporates the Maxim ModelGauge m5 algorithm that combines the excellent short-term accuracy and linearity of a coulomb counter with the excellent longterm stability of a voltage-based fuel gauge, along with temperature compensation to provide industry-leading fuel-gauge accuracy. The device also includes Cycle+ charger control, improved aged adaptation, improved state-of-charge (SOC) accuracy to empty, and increased temperature measurement accuracy. ModelGauge m5 cancels offset accumulation error in the coulomb counter. while providing better short-term accuracy than any purely voltage-based fuel gauge. Additionally, the ModelGauge m5 algorithm does not suffer from abrupt corrections that normally occur in coulomb-counter algorithms, since tiny continual corrections are distributed over time. The device automatically compensates for aging, temperature, and discharge rate and provides accurate SOC in mAh or % over a wide range of operating conditions. The device provides two methods for reporting the age of the battery: reduction in capacity and cycle odometer. The device provides precision measurements of current, voltage, and temperature. Temperature of the battery pack is measured using an external thermistor supported by ratiometric measurements on an auxiliary input. A 2-wire (I²C) interface provides access to data and control registers.

ModelGauge m5 Algorithm

The ModelGauge m5 algorithm combines a high-accuracy coulomb counter with a voltage fuel gauge (VFG), as described in Figure 17. Classical coulomb-counterbased fuel gauges have excellent linearity and short-term performance. However, they suffer from drift due to the accumulation of the offset error in the current-sense measurement. Although the offset error is often very small, it cannot be eliminated, causes the reported capacity error to increase over time, and requires periodic corrections.

Corrections are usually performed at full or empty. Some other systems also use the relaxed battery voltage to perform corrections. These systems determine the SOC based on the battery voltage after a long time of no current flow. Both have the same limitation: if the correction condition is not observed over time in the actual application. the error in the system is boundless. The performance of classic coulomb counters is dominated by the accuracy of such corrections. Classical voltage-measurement-based SOC estimation has poor accuracy due to inadequate cell modeling, but does not accumulate offset error over time. The device includes an advanced VFG, which estimates open-circuit voltage (OCV), even during current flow, and simulates the nonlinear internal dynamics of a lithium-ion (Li+) battery to determine the SOC with improved accuracy. The model considers the time effects of a battery caused by the chemical reactions and impedance in the battery to determine SOC based on table lookup. This SOC estimation does not accumulate offset error over time. The ModelGauge m5 algorithm combines a highaccuracy coulomb counter with a VFG. The complementary combined result eliminates the weaknesses of both the coulomb counter and the VFG, while providing the strengths of both. A mixing algorithm combines the VFG capacity with the coulomb counter and weighs each result so that both are used optimally to determine the battery state. In this way, the VFG capacity result is used to continuously make small adjustments to the battery state, canceling the coulomb-counter drift. The ModelGauge m5 algorithm uses this battery state information and accounts for temperature, battery current, age, and application parameters to determine the remaining capacity available to the system. The ModelGauge m5 algorithm continually adapts to the cell and application through independent learning routines. As the cell ages, its change in capacity is monitored and updated, and the VFG dynamics adapt based on cell voltage behavior in the application.

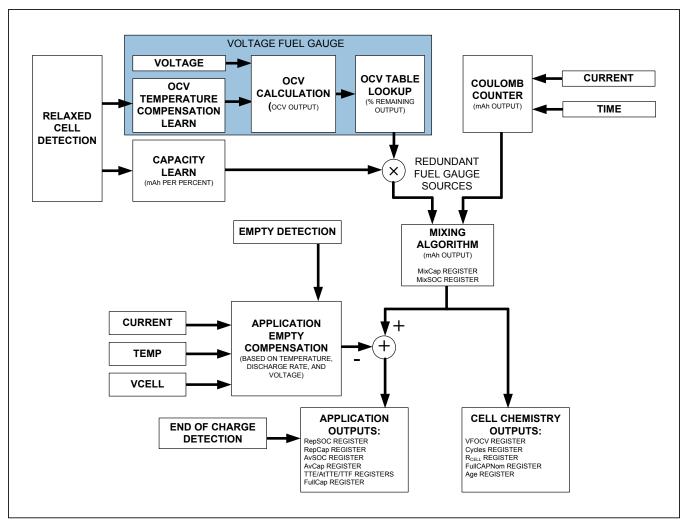


Figure 17. ModelGauge m5 Block Diagram

OCV Estimation and Coulomb-Count Mixing

The core of the ModelGauge m5 algorithm is a mixing algorithm that combines the OCV state estimation with the coulomb counter. After power-on reset of the IC, coulomb-count accuracy is unknown. The OCV state estimation is weighted heavily compared to the coulomb-count output. As the cell progresses through cycles in the application, coulomb-counter accuracy improves and the mixing algorithm alters the weighting so that the coulomb-counter result is dominant. From this point forward, the IC switches to servo mixing. Servo mixing provides a fixed

magnitude continuous error correction to the coulomb count, up or down, based on the direction of error from the OCV estimation. This allows differences between the coulomb count and OCV estimation to be corrected quickly. The resulting output from the mixing algorithm does not suffer drift from current measurement offset error and is more stable than a stand-alone OCV estimation algorithm. Initial accuracy depends on the relaxation state of the cell. The highest initial accuracy is achieved with a fully relaxed cell.

Fuel-Gauge Empty Compensation

As the temperature and discharge rate of an application changes, the amount of charge available to the application also changes. The ModelGauge m5 algorithm distinguishes between remaining capacity of the cell (RemCapMIX), and remaining capacity of the application (RemCapAV), and reports both results to the user.

Fuel-Gauge Learning

The device periodically makes internal adjustments to cell characterization and application information to remove initial error and maintain accuracy as the cell ages. These adjustments always occur as small under corrections to prevent instability of the system and prevent any noticeable jumps in the fuel-gauge outputs. Learning occurs automatically without any input from the host. To maintain learned accuracy through power loss, the host must periodically save learned information and then restore after power is returned. See the <u>Power-Up and Power-On Reset</u> section for details:

- Full Capacity Available to Application (FullCAP):
 This is the total capacity available to the application at full. FullCAP is updated near the end of charging when termination is detected. See the End-of-Charge Detection section.
- Cell Capacity (FullCapNom): This is the total cell capacity at full, according to the VFG. This includes some capacity that is not available to the application at high loads and/or low temperature. The device periodically compares percent change based on OCV measurement vs. coulomb-count change as the cell charges and discharges. This information allows the device to maintain an accurate estimation of the cell's capacity in mAh as the cell ages.
- Voltage Fuel-Gauge Adaptation: The device observes the battery's relaxation response and adjusts the dynamics of the VFG. This adaptation adjusts the RCOMP0 register during qualified cell relaxation events.
- Empty Compensation: The device updates internal data whenever cell empty is detected (VCELL < V_ empty) to account for cell age or other cell deviations from the characterization information.

Determining Fuel-Gauge Accuracy

To determine the true accuracy of a fuel gauge, as experienced by end users, the battery should be exercised in a dynamic manner. The end-user accuracy cannot be understood with only simple cycles. To challenge a correction-based fuel gauge, such as a coulomb counter, test the battery with partial loading sessions. For example, a typical user can operate the device for 10 minutes and then stop use for an hour or more. A robust test method includes these kinds of sessions many times at various loads, temperatures, and duration.

Initial Accuracy

The device uses the first voltage reading after power-up or after cell insertion to determine the starting output of the fuel gauge. It is assumed that the cell is fully relaxed prior to this reading; however, this is not always the case. If the cell was recently charged or discharged, the voltage measured by the device may not represent the true state of charge of the cell, resulting in initial error in the fuel-gauge outputs. In most cases, this error is minor and is quickly removed by the fuel-gauge algorithm during normal operation.

Cycle+ Charger Control

ModelGauge m5 is paired internally with a programmable, high voltage, high current charger. The charger has adjustable JEITA thresholds and works with the Cycle+scheme to control the charge current vs. state-of-charge. In addition, Cycle+ provides smart-full charge termination, which can control charger charge-termination according to the OCV prediction of the fuel gauge. This helps to support more accurate charge termination and is less sensitive to system load transients at the end of charging.

End-of-Charge Detection

The device detects the end of a charge cycle when the application current falls into the band set by the ICHGTerm register value. By monitoring both the current and average current registers, the device can reject false end-of-charge events such as application load spikes or early charge-source removal. When a proper end-of-charge event is detected, the device learns a new FullCAP register value based on the RemCapREP output. If the old FullCAP value was too high, it is adjusted downward after the last valid end-of-charge detection. If the old FullCAP was too low, it is adjusted upward to match RemCapREP. This prevents the calculated state of charge from ever reporting a value greater than 100%.

Power-Up and Power-On Reset

Any power-on reset (POR) of the device resets all memory locations to their default POR value. This removes any custom cell characterization and application data, affects ALRT interrupt and shutdown mode settings, and resets all learned adjustments made by the fuel gauge. To maintain accuracy of the fuel gauge and reset operation settings of the device, the host must reload all application memory data and restore all learned fuel gauge information. Note that the device can take up to 445ms to completely reset operation after a POR event occurs. Saved data should not be restored until after this period is over. The following procedure is recommended:

- Read Status register. If POR = 0, exit.
- Wait 600ms for POR operation to fully complete.
- · Restore all application register values.
- Restore fuel gauge learned-value information (see the Save and Restore Registers section).
- · Clear POR bit.

Save and Restore Registers

The device is designed to operate outside the battery pack and can therefore be exposed to power loss when in the application. To prevent the loss of learned information during power cycles, a save-and-restore procedure can be used to maintain register values in nonvolatile memory external to the device. The registers (see Table 1) must be stored externally and then rewritten to the device after power-up to maintain a learned state of operation. Note that some registers are application outputs, some registers are for internal calculations, and some are characterization setup registers. Registers that are not internal are described in their own sections. These values should be stored by the application at periodic intervals. Some recommended back-up events are:

- End-of-charge
- End-of-discharge
- · Prior to application entering shutdown state

The host is responsible for loading the default characterization data at first power-up of the device, and restoring the default characterization data, plus learned information on subsequent power-up events.

Battery Removal and Insertion

The device detects when a cell has been removed or inserted into the application. This allows the device to

adjust to the new cell to maintain accuracy. The removal detection feature also allows the device to quickly warn the host processor through interrupt of impending power loss if enabled. Detection occurs by monitoring the AIN pin voltage compared to the THRM pin. Whenever a cell is present, the external resistor-divider network sets the voltage of AIN. When the cell is removed, the remaining external resistor pulls AIN to the THRM pin voltage level. Whenever VAIN < VTHRM - VDETF, the device determines that a cell is present in the application. If VAIN > VTHRM - VDETR, the device determines that no cell is present at that time.

Cell Insertion (IC Already Powered)

The device is ready to detect a cell insertion if either the ETHRM or FTHRM bits of the CONFIG register are set to enable the THRM pin output. When a cell insertion is detected, the fuel gauge is reset and all fuel-gauge outputs are updated to reflect the SOC of the newly inserted cell. This process can take up to 1.845s (FTHRM = 0) or 620ms (FTHRM = 1) from time of insertion. Note that the device uses the cell voltage as a starting point for the fuel gauge. If the cell voltage is not fully relaxed at time of insertion, the fuel gauge begins with some initial error. See the Fuel-Gauge Learning section for details. The host can disable this feature by clearing the enBi1 bit in the MiscCFG register. The device can also be configured to alert the host when cell insertion occurs. When Bei = 1 in the CONFIG register, the device generates an interrupt on the ALRT pin at the start of the first temperature conversion after insertion. This could take up to 1.4s to occur. This feature is useful if the application uses more than one cell type and the device must be reconfigured at each insertion.

Cell Removal

The device detects a cell removal if either the ETHRM or FTHRM bits of the CONFIG register are set to enable the THRM pin output. Cell removal does not affect IC operation. The device continues to update fuel-gauge outputs. The host should monitor the Br and Bst bits of the Status register to determine if the fuel-gauge outputs are valid. The device can also be configured to alert the host when cell removal occurs. When Ber = 1 in the CONFIG register, the device generates an interrupt on the ALRT pin at the start of the first temperature conversion after removal. This could take up to 1.4s to occur. This feature is useful if the application uses more than one cell type and the device must be reconfigured at each insertion.

Fast Detection of Cell Removal

The device can be configured to quickly alert the host of impending power loss on cell removal. This fast response allows the system to quickly and gracefully hibernate to prevent power loss during battery swap. When Ber = 1, FTHRM = 1, and ALRTp = 0 in the CONFIG register, an interrupt on the ALRT pin is generated within 100Fs after VAIN becomes greater than VTHRM - VDETR. If fast detection is used, it is recommended that all other IC interrupts are disabled to prevent the host from spending time determining the cause of the interrupt. Fast detection of cell removal has no effect on fuel-gauge operation, but leaving the external resistor-divider active increases current consumption of the application.

Modes of Operation

The device operates in one of two power modes: active and shutdown. While in active mode, the device operates as a high-precision battery monitor with temperature, voltage, auxiliary inputs, current, and accumulated current measurements acquired continuously, and the resulting values updated in the measurement registers. READ and WRITE access is allowed only in active mode. In shutdown mode, the LDO is disabled and all activity stops, although volatile RAM contents remain preserved. All A/D register and fuel-gauge output values are maintained. There are several options for entering shutdown:

Entering shutdown:

- SHUTDOWN Command—Write the CONFIG register SHDN = 1 through the I²C interface; wait for longer than the SHDNTIMER register value.
- Pack Removal—Pack removal detection is valid for longer than the SHDNTIMER register value and the CONFIG register AINSH = 1.
- I²C Shutdown—I²C lines both persist low for longer than the SHDNTIMER register value and the CON-FIG register I²CSH = 1.
- ALRT Shutdown—Shutdown occurs when the
 ALRT line is externally driven low for longer than the
 SHDNTIMER register value (ALSH = 1 and ALRTp =
 0), or the ALRT line is externally driven high for longer
 than the SHDNTIMER register value (ALSH = 1 and
 ALRTp = 1). See the CONFIG Register (1Dh) section.

These shutdown entry modes are all programmable according to application. Shutdown events are gated by the SHDNTIMER register, which allows a long delay between the shutdown event and the actual shutdown. By behaving this way, the device takes the best reading of the relaxation voltage.

Exiting shutdown:

- I²C Wakeup—Any edge on SCL/SDA.
- ALRT Wakeup—Any edge on ALRT line and (ALSH = 1 or I2CSH = ALSH = 0).
- Reset—IC is power cycled.

See the <u>Status and Configuration Registers</u> section for detailed descriptions of the SHDNTIMER and CONFIG registers. The state of the device when returning to active mode differs depending on the triggering event. Host software can monitor the POR and Bi status bits to determine what type of event has occurred.

ALRT Function

The Alert Threshold registers allow interrupts to be generated by detecting a high or low voltage, a high or low temperature, or a high or low SOC. Interrupts are generated on the ALRT pin open-drain output driver. An external pullup is required to generate a logic-high signal. Note that if the pin is configured to be logic-low when inactive, the external pullup increases current drain. The ALRTp bit in the CONFIG register sets the polarity of the ALRT pin output. Alerts can be triggered by any of the following conditions:

- **Battery Removal**—(VAIN > VTHRM VDETR) and battery removal detection enabled (Ber = 1).
- Battery Insertion—(VAIN < VTHRM VDETF) and battery insertion detection enabled (Bei = 1).
- Over/Under Voltage—VALRT threshold violation (upper or lower) and alerts enabled (Aen = 1).
- Over/Under Temperature—TALRT threshold violation (upper or lower) and alerts enabled (Aen = 1).
- Over/Under SOC—SALRT threshold violation (upper or lower) and alerts enabled (Aen = 1).
- **dSOC** 1% change in SOC (Config.dSOC = 1)
- Charger Communication Failure—Set when FG fails to communicate with the charger. Controlled by Config.FCFE and Config.ICFE.

To prevent false interrupts, the threshold registers should be initialized before setting the Aen bit. Alerts generated by battery insertion or removal can only be reset by clearing the corresponding bit in the status register. Alerts generated by a threshold-level violation can be configured to be cleared only by software, or cleared automatically when the threshold level is no longer violated. See the CONFIG (1Dh) register description for details of the alert function configuration.

IC Memory Map

The device has a 256-word linear memory space containing all user-accessible registers. All registers are 16 bits wide and are read and written as 2-byte values. When the MSB of a register is read, the MSB and LSB are latched simultaneously and held for the duration of the Read Data command. This prevents updates to the LSB during the read, ensuring synchronization between the 2 register bytes. All locations are volatile RAM and lose their data in the event of power loss. Data is retained during device shutdown. Each register has a power-on-reset value that

it defaults to at power-up. Word addresses designated as reserved return an undetermined when read. These locations should not be written.

ModelGauge m5 Registers

All ModelGauge m5 registers are shown in $\underline{\text{Table }}$ 1. The register address is composed by two HEX digits shown given by first row and first column. All the registers are grouped by their functions and described in the $\underline{\textit{ModelGauge m5}}$ $\underline{\textit{Output Registers}}$ section.

Table 1. ModelGauge m5 Registers

0X6C I ² C COMMAND ADDRESS	0X0_	0X1_	0X2_	0X3_	0X4_	0X5_ 0X7_	0X8_	0X9_	0XA_	0XB_	0XC_	0XD_	0XE_	0XF_
0x0_0	STATUS	FullCAP	TTF	Reserved	Reserved					Status2		Reserved		Reserved
0x0_1	VALRT_Th	TTE	DevName	Reserved	Reserved					Reserved		ChargeState0		Reserved
0x0_2	TALRT_Th	QRtable00	QRtable10	QRtable20	QRtable30					TALRT_Th2		ChargeState1		Reserved
0x0_3	SALRT_Th	FullSOCthr	FullCAPNom	AtTTF	Reserved					Reserved		ChargeState2		Reserved
0x0_4	AtRate	RSLOW	TempNom	Reserved	Reserved					Reserved		ChargeState3		Reserved
0x0_5	RepCap	Reserved	TempLim	FullCapRep	dQ_acc					TTF_CFG		ChargeState4		Reserved
0x0_6	RepSOC	AvgTA	Reserved	lavg_empty	dP_acc]			l _	CV_MixCap		ChargeState5		Reserved
0x0_7	Age	Cycles	AIN0	FCTC	Reserved	Res			RCOMP	CV_HalfTime	Res	ChargeState6	Res	Reserved
0x0_8	TEMP	DesignCap	LearnCFG	RCOMP0	Reserved	Reserved	OCV	CAP	MP s	CGTempCo	Reserved	ChargeState7	Reserve	Reserved
0x0_9	VCELL	AvgVCELL	FilterCFG	TempCo	ConvgCfg				seg	Curve	۵	JEITA_Volt	٥	Reserved
0x0_A	Current	MaxMinTemp	RelaxCFG	V_empty	VFRemCap					Reserved		JEITA_Curr		Reserved
0x0_B	AvgCurrent	MaxMinVolt	MiscCFG	Reserved	Reserved]				Config2		SmartChgCfg]	VFOCV
0x0_C	Qresidual	MaxMinCurr	TGAIN	Reserved	Reserved					Vripple		AtQresidual]	Reserved
0x0_D	MixSOC	CONFIG	TOFF	Reserved	QH	1				RippleCfg		AtTTE		Reserved
0x0_E	AvSOC	ICHGTerm	CGAIN	TIMER	Reserved	1				TIMERH		AtAvSOC		Reserved
0x0_F	MixCap	AvCap	COFF	SHDNTIMER	Reserved					MaxError		AtAvCap		VFSOC

System Protections

V_SY_S Undervoltage Lockout (V_SY_SUV_LO)

When SYS voltage falls below $V_{SYSUVLO}$, the device's type O registers are reset.

When the charger input is valid and battery is present and:

VPQDB<VSYS<VSYS UVLO

Q_{BAT} is on and SYS is shorted to BATT.

When:

$$0 < V_{SYS} < V_{PQDB}$$

 $Q_{\mbox{\footnotesize{BAT}}}$ is off, but the charger pulls up SYS from BAT with a constant current of 50mA.

When charger input is invalid and battery is present and:

VPQDB<VSYS<VSYSUVLO

QBAT is on and SYS is shorted to BAT.

When:

$$0 < V_{SYS} < V_{PQDB}$$

QRAT is off.

V_{SYS} Overvoltage Lockout (V_{SYSOVLO})

Ideally, V_{SYS} should not exceed the battery charge termination threshold. Systems must be designed so that V_{SYS} never exceeds 4.8V (transient and steady-state). If the V_{SYS} exceeds $V_{SYSOVLO}$ during a fault, the MAX77818 resets the charger and fuel gauge O type registers.

Safeout LDO

The MAX77818 integrates two high voltage input LDOs, with I²C programmable on/off control, active discharge and programmable output voltages, ideal for powering low voltage rated USB systems. Refer to the register 0xC6: SAFEOUT LDO Control detailed description. SAFEOUT1 is enabled by default once charger detection is completed and CHGIN is valid. SAFEOUT1 is off by default.

I²C Interface

The MAX77818 acts as a slave transmitter/receiver. The MAX77818 has the following slave addresses:

Slave Addresses

Charger: 0xD2/D3h

Clogic, GTEST and Safeout LDOs: 0xCCh/0xCDh

Fuel Gauge: 0x6C/0x6D. See the Fuel Gauge I²C Protocol for details in the *Fuel Gauge Electrical Characteristics* section.

I²C Bit Transfer

One data bit is transferred for each clock pulse. The data on SDA must remain stable during the high portion of the clock pulse as changes in data during this time are interpreted as a control signal.

I²C Start and Stop Conditions

Both SDA and SCL remain high when the bus is not busy. A high-to-low transition of SDA, while SCL is high is defined as the start (S) condition. A low-to-high transition of SDA while SCL is high is defined as the stop (P) condition.

I²C System Configuration

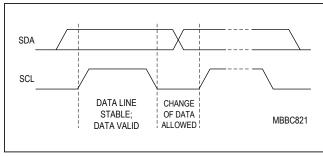
A device on the I²C bus that generates a message is called a transmitter and a device that receives the message is a receiver. The device that controls the message is the master. The devices that are controlled by the master are called slaves.

I²C Acknowledge

The number of data bytes between the start and stop conditions for the transmitter and receiver are unlimited.

Each 8-bit byte is followed by an acknowledge bit. The acknowledge bit is a high-level signal put on SDA by the transmitter during the time the master generates an extra acknowledge-related clock pulse. A slave receiver that is addressed must generate an acknowledge after each byte it receives. Also, a master receiver must generate an acknowledge after each byte it receives that has been clocked out of the slave transmitter.

The device that acknowledges must pull down the SDA line during the acknowledge clock pulse, so that the SDA line is stable low during the high period of the acknowledge clock pulse (set-up and hold times must also be met). A master receiver must signal an end of data to the transmitter by not generating an acknowledge on the last byte that has been clocked out of the slave. In this case, the transmitter must leave SDA high to enable the master to generate a stop condition.



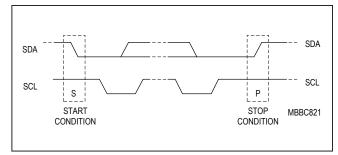


Figure 18. I²C Bit Transfer

Figure 19. I²C Start and Stop

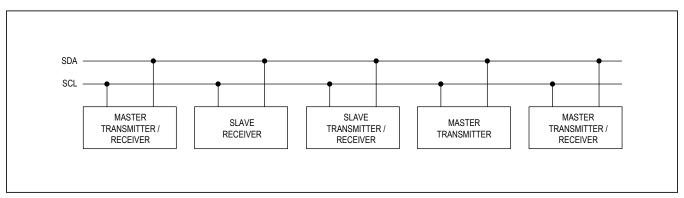


Figure 20. I²C System Configuration

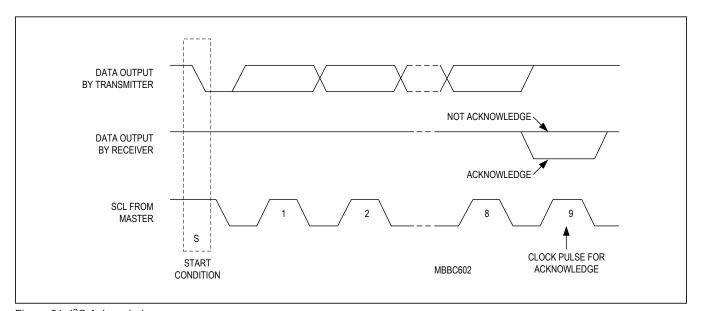
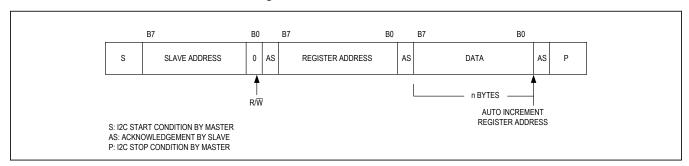


Figure 21. I²C Acknowledge

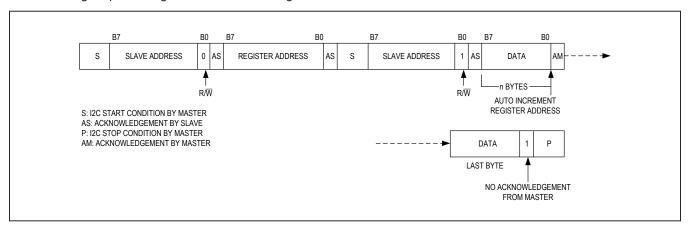
Master Transmits (Write Mode)

When master writes to slave, use the following format:



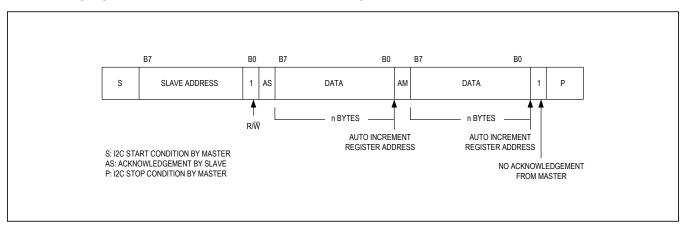
Master Reads After Setting Register Address (Write Register Address and Read Data)

When reading a specific register, use the following format:



Master Reads Register Data Without Setting Register Address (Read Mode)

When reading registers from the first address, use the following format:



I²C Register Map and Detailed Descriptions

INTB

The INTB is the hardware pin used to notify the host processor that one or more of the unmasked interrupt bits toggled. Masking the interrupt by setting INT_MASK_ bit to '1' prevents the INTB from asserting. The INTB pin deasserts (pulled high by off-chip pullup) as soon as the read sequence of the last INT_ register that contains an active interrupt starts. FG interrupts are cleared by setting new threshold values.

The application processor reads the interrupts in two steps. First, the AP reads the INTSRC register. This is a read-only register that indicates which functional block is generating the interrupt (i.e., charger and FG). Depending on the result of the read, the next step is to read the actual interrupt registers pertaining to the functional block.

For example, if the application processor reads 0x02 from register 0x22: INTSRC, it means the top-level MAX77818 block has an interrupt generated. The next step is to read register 0x24: SYSTEM Interrupt in order to understand the source of the interrupt.

Register Type and Reset Conditions

Type S: Registers are reset each time when SYS < SYS POR (~1.55V).

Type O: Registers are reset each time when SYS < SYS UVLO (2.55V max) or SYS > SYS OVLO or Die temp > 165° (or MAX77818 transitions from on to off state).

Top Level I²C Registers

The MAX77818 acts as a slave transmitter/receiver. The slave address of the MAX77818 top is 0xCCh/0xCDh (OTP option for 0xDC/0xDDh). The least significant bit is the read/write indicator.

The MAX77818's tope level has the following registers:

0x20: PMIC ID Register

NAME	FUNCTION	ADDR	TYPE	RESET
PMICID	PMIC ID	0x20	0	0x23

BIT	MODE	NAME	RESET	DESCRIPTION
3:0	R	ID	0011	ID of MAX77818
7:4	R	ID	0010	ID UI IVIAA77616

0x21: PMIC Version/Rev Register

NAME	FUNCTION	ADDR	TYPE	RESET
PMICREV	PMIC revision	0x21	0	0x80

BIT	MODE	NAME	RESET	DESCRIPTION
				Pass
2:0	R	REV		0b000 = pass 1
2.0	K	KEV	_	0b001 = pass 2
				0b010 = pass 3
		VERSION	_	Version
				0b00000 = zmo
				0b00001 = ymo
7:3	R			
				Null Trim Version
				0b10000 = tmo
				0b10001 = umo

0x22: Interrupt Source

NAME	FUNCTION	ADDR	TYPE	RESET
INTSRC	Interrupt source	0x22	S	0x00

BIT	MODE	NAME	RESET	DESCRIPTION
0	R CHGR INT		0	0: No charger interrupt.
		0.1011		1: Charger interrupt is detected.
1	R	FG INT	0	0 = No interrupt pending from FG block.
		10_1111		1 = interrupt from FG block is detected .
2	R	SYS_INT	0	0: No SYS INT
				1: SYS interrupt is detected.
3	R	RSVD	0	Reserved
4	R	RSVD	0	Reserved
5	R	RSVD	0	Reserved
6	R	RSVD	0	Reserved
7	R	RSVD	0	Reserved

0x23: Interrupt Source Mask

NAME	FUNCTION	ADDR	TYPE	RESET
INTSRCMASK	Interrupt source mask	0x23	S	0xFF

BIT	MODE	NAME	RESET	DESCRIPTION
0	R/W	CHGR_INT_MASK	1	1: Charger interrupt is masked.
1	R/W	FG_INT_MASK	1	1: FG interrupt is masked.
2	R/W	SYS_INT_MASK	1	1: SYS interrupt is masked.
3	R/W	RSVD	1	Reserved
4	R/W	RSVD	1	Reserved
5	R/W	RSVD	1	Reserved
6	R/W	RSVD	1	Reserved
7	R/W	RSVD	1	Reserved

0x24: SYSTEM Interrupt

NAME	FUNCTION	ADDR	TYPE	RESET
SYSINTSRC	SYS interrupt source	0x24	S	0x00

BIT	MODE	NAME	RESET	DESCRIPTION
0	R	SYSUVLO_INT	0	0: No SYSUVLO INT. 1: SYSUVLO interrupt is detected (falling).
1	R	SYSOVLO_INT	0	No SYSOVLO interrupt. SYSOVLO interrupt is detected (rising and falling).
2	R	TSHDN_INT		0: No TSHDN interrupt. 1: TSHDN interrupt is detected.
3	R	RSVD	0	Reserved
4	R	RSVD	0	Reserved
5	R	RSVD	0	Reserved
6	R	RSVD	0	Reserved
7	R	TM_INT	0	O: Test mode interrupt is not set. 1: Test mode interrupt is set.

0x26: SYSTEM Interrupt Source Mask

NAME	FUNCTION	ADDR	TYPE	RESET
SYSINTMASK	System interrupt mask	0x26	S	0xFF

BIT	MODE	NAME	RESET	DESCRIPTION
0	R/W	SYSUVLO_INT_MASK	1	1: SYSUVLO interrupt is masked.
1	R/W	SYSOVLO_INT_MASK	1	1: SYSUVLO interrupt is masked.
2	R/W	TSHDN_INT_MASK	1	1: Thermal shutdown interrupt is masked.
3	R/W	RSVD	1	Reserved
4	R/W	RSVD	1	Reserved
5	R/W	RSVD	1	Reserved
6	R/W	RSVD	1	Reserved
7	R/W	TM_INT_MASK	0	1: INT test mode interrupt is masked.

0xC6: SAFEOUT LDO Control

NAME	FUNCTION	ADDR	TYPE	RESET
SAFEOUTCTRL	SAFEOUT linear regulator control	0xC6	0	0x75

BIT	MODE	NAME	RESET	DESCRIPTION
1:0	R/W	SAFEOUT1[1:0]	01	SAFEOUT1 output voltage 00: 4.85V 01: 4.90V (default) 10: 4.95V 11: 3.3V.
3:2	R/W	SAFEOUT2[3:2]	01	SAFEOUT2 output voltage 00: 4.85V 01: 4.90V (default) 10: 4.95V 11: 3.3V
4	R/W	ACTDISSAFEO1	1	0: No active discharge 1: Active discharge
5	R/W	ACTDISSAFEO2	1	0: No active discharge 1: Active discharge
6	R/W	ENSAFEOUT1	1	SAFEOUTLDO1 enable bit 0: Disable SAFEOUT1. 1: Enable SAFEOUT1.
7	R/W	ENSAFEOUT2	0	SAFEOUTLDO2 enable bit 0: Disable SAFEOUT2. 1: Enable SAFEOUT2.

Charger I²C Registers

The MAX77818's charger has convenient default register settings and a complete charger state machine that allow it to be used with minimal software interaction. Software interaction with the register map enhances the charger by allowing a high degree of configurability. An easy-to-navigate interrupt structure and in-depth status reporting allows software to quickly track the changes in the charger's status.

Register Protection

The CHG_CNFG_01, CHG_CNFG_02, CHG_CNFG_03, CHG_CNFG_04, CHG_CNFG_05, and CHG_CNFG07 registers contain settings for static parameters that are associated with a particular system and battery. These static settings are typically set once each time the system's microprocessor runs its boot-up initialization code; then they are not changed again until the microprocessor reboots. CHGPROT allows for blocking the "write" access to these static settings to protect them from being changed unintentionally. This protection is particularly useful for critical parameters such as the battery charge current CHG_CC and the battery charge voltage CHG_CV_PRM.

Determine the following registers bit settings by considering the characteristics of the battery. Maxim recommends that CHG_CC be set to the maximum acceptable charge rate for your battery – there is typically no need to actively adjust the CHG_CC setting based on the capabilities of the source at CHGIN, system load, or thermal limitations of the PCB; the Smart Power Selector intelligently manages all these parameters to optimize the power distribution.

Charger Restart Threshold → CHG_RSTRT
Fast-Charge Timer (t_{FC}) → FCHGTIME
Fast-Charge Current → CHG CC

Top-Off Time → TO_TIME
Top-Off Current → TO_ITH

Battery Regulation Voltage → CHG CV PRM

Determine the following register bit settings by considering the characteristics of the system:

Low Battery Prequalification Enable → PQEN
Minimum System Regulation Voltage → MINVSYS
Junction Temperature Thermal Regulation
Loop Setpoint → REGTEMP

Interrupt, Mask, OK, and Detail Registers

The battery charger section of the MAX77818 provides detailed interrupt generation and status for the following subblocks:

- Charger Input
- Charger State Machine
- Battery
- Bypass Node

State changes on any subblock report interrupts through the CHG_INT register. Interrupt sources are masked from affecting the hardware interrupt pin when bits in the CHG_INT_MASK register are set. The CHG_INT_OK register provides a single-bit status indication of whether the interrupt generating subblock is okay or not. The full status of interrupt generating subblock is provided in the CHG_DETAILS_00, CHG_DETAILS_01, CHG_DETAILS_02, and CHG_DETAILS_03 registers.

Note that CHG_INT, CHG_INT_MASK, and CHG_INT_ OK use the same bit position for each interrupt generating block to simplify software development.

Interrupt bits are automatically cleared upon reading a given interrupt register. When all pending CHG_INT interrupts are cleared, the top level interrupt bit deasserts.

CHG_INT Register Bit Description (0xB0)

NAME	FUNCTION	ADDR	TYPE	RESET
CHG_INT	Charger interrupt	0xB0	0	0x00

BIT	MODE	NAME	RESET	DESCRIPTION
0	R/C	BYP_I	0	Bypass Node Interrupt 0 = The BYP_OK bit has not changed since the last time this bit was read. 1 = The BYP_OK bit has changed since the last time this bit was read.
1	R/C	RSVD	0	Reserved
2	R/C	BATP_I	0	Battery Presence Interrupt. 0 = The BATP_OK bit has not changed since the last time this bit was read. 1 = The BATP_OK bit has changed since the last time this bit was read.s
3	R/C	BAT_I	0	Battery Interrupt 0 = The BAT_OK bit has not changed since the last time this bit was read. 1 = The BAT_OK bit has changed since the last time this bit was read.
4	R/C	CHG_I	0	Charger Interrupt 0 = The CHG_OK bit has not changed since the last time this bit was read. 1 = The CHG_OK bit has changed since the last time this bit was read.
5	R/C	WCIN_I	0	WCIN Interrupt. 0 = The WCIN_OK bit has not changed since the last time this bit was read. 1 = The WCIN_OK bit has changed since the last time this bit was read.
6	R/C	CHGIN_I	0	CHGIN Interrupt. 0 = The CHGIN_OK bit has not changed since the last time this bit was read. 1 = The CHGIN_OK bit has changed since the last time this bit was read.
7	R/C	AICL_I	0	AICL interrupt 0=The AICL_OK bit has not changed since the last time this bit was read. 1=The AICL_OK bit has changed since the last time this bit was read.

CHG_INT_MASK Register Bit Description (0xB1)

NAME	FUNCTION	ADDR	ТҮРЕ	RESET
CHG_INT_MASH	Charger interrupt mask	0xB1	0	0xFF

BIT	MODE	NAME	RESET	DESCRIPTION
0	R/W	BYP_M	1	Bypass Interrupt Mask 0 = Unmasked 1 = Masked
1	R/W	RSVD	1	Reserved
2	R/W	BATP_M	1	Battery Presence Interrupt Mask 0 = Unmasked 1 = Masked
3	R/W	BAT_M	1	Battery Interrupt Mask 0 = Unmasked 1 = Masked
4	R/W	CHG_M	1	Charger Interrupt Mask 0 = Unmasked 1 = Masked
5	R/W	WCIN_M	1	WCIN Interrupt Mask 0 = Unmasked 1 = Masked
6	R/W	CHGIN_M	1	CHGIN Interrupt Mask 0 = Unmasked 1 = Masked
7	R/W	AICL_M	1	AICL Interrupt Mask 0 = Unmasked 1 = Masked

CHG_INT_OK Register Bit Description (0xB2)

NAME	FUNCTION	ADDR	ТҮРЕ	RESET
CHG_INT_OK	Charger status	0xB2	0	0x00

BIT	MODE	NAME	RESET	DESCRIPTION
0	R	BYP_OK	0	Single-Bit Bypass Status Indicator. See BYP_DTLS for more information. 0 = Something powered by the bypass node has hit current limit. i.e., BYP_DTLS ≠ 0x00. 1 = The bypass node is okay. i.e., BYP_DTLS = 0x00.
1	R	RSVD	0	Reserved
2	R	BATP_OK	0	BAT present status indicator. 0 = Main Battery is not present 1 = Main Battery is present.
3	R	BAT_OK	0	Single-Bit Battery Status Indicator. See BAT_DTLS for more information. 0 = The battery has an issue or the charger has been suspended, i.e., BAT_DTLS ≠ 0x03 or 0x04 1 = The battery is okay. i.e., BAT_DTLS = 0x03 or 0x04
4	R	снд_ок	0	Single-Bit Charger Status Indicator. See CHG_DTLS for more information. 0 = The charger has suspended charging or TREG = 1 i.e., CHG_DTLS ≠ 0x00 or 0x01 or 0x02 or 0x03 or 0x05 or 0x08 1 = The charger is okay or the charger is off i.e., CHG_DTLS = 0x00 or 0x01 or 0x02 or 0x03 or 0x05 or 0x08
5	R	WCIN_OK	0	Single-Bit WCIN Input Status Indicator. See WCIN_DTLS for more information. 0 = The WCIN input is invalid. i.e., WCIN_DTLS ≠ 0x03. 1 = The WCIN input is valid. i.e., WCIN_DTLS = 0x03.
6	R	CHGIN_OK	0	Single-Bit CHGIN Input Status Indicator. See CHGIN_DTLS for more information. 0 = The CHGIN input is invalid. i.e., CHGIN_DTLS ≠ 0x03. 1 = The CHGIN input is valid. i.e., CHGIN_DTLS = 0x03.
7	R	AICL_OK	0	AICL_OK 0 = AICL mode 1 = Not in AICL mode

CHG_DETAILS_00 Register Bit Description (0xB3)

NAME	FUNCTION	ADDR	ТҮРЕ	RESET
CHG_DTLS_00	Charger details 00	0xB3	О	0x00

BIT	MODE	NAME	RESET	DESCRIPTION
0	R	BATP_DTLS	0	Battery Detection 0 = Battery presence 1 = No battery presence
2:1	R	RSVD	00	Reserved
4:3	R	WCIN_DTLS	00	WCIN Details 0x00 = VWCIN is invalid. V _{WCIN} < V _{WCIN_UVLO} 0x01 = VWCIN is invalid. V _{WCIN} < V _{MBAT} + V _{WCIN2SYS} and V _{WCIN} > V _{WCIN_UVLO} 0 x 02 = VWCIN is invalid. V _{WCIN} >V _{WCIN_OVLO} 0 x 03 = VWCIN is valid. V _{WCIN} > V _{WCIN_UVLO} , VWCIN > V _{MBAT} + V _{WCIN2SYS} , V _{WCIN_OVLO}
6:5	R	CHGIN_DTLS	00	CHGIN Details 0x00 = VBUS is invalid. V _{CHGIN} < V _{CHGIN} _UVLO 0x01 = VBUS is invalid. V _{CHGIN} < V _{MBAT} + V _{CHGIN2SYS} and V _{CHGIN} > V _{CHGIN} _UVLO 0x02 = VBUS is invalid. V _{CHGIN} > V _{CHGIN} _OVLO 0x03 = VBUS is valid. V _{CHGIN} > V _{CHGIN} _UVLO, V _{CHGIN} > V _{MBAT} + V _{CHGIN2SYS} , V _{CHGIN} < V _{CHGIN} _OVLO
7	R	RSVD	0	Reserved

CHG_DETAILS_01 Register Bit Description (0xB4)

NAME	FUNCTION	ADDR	TYPE	RESET
CHG_DTLS_01	Charger details 01	0xB4	0	0x00

BIT	MODE	NAME	RESET	DESCRIPTION
3:0	R	CHG_DTLS	0000	Charger Details 0x00 = Charger is in dead-battery prequalification or low-battery prequalification mode, CHG_OK = 1, V _{MBATT} < V _{PQLB} , T _J < T _{JSHDN} 0x01 = Charger is in fast-charge constant current mode, CHG_OK = 1, V _{MBATT} < V _{BATREG} , T _J < T _{JSHDN} 0x02 = Charger is in fast-charge constant voltage mode, CHG_OK = 1, V _{MBATT} = V _{BATREG} , T _J < T _{JSHDN} 0x03 = Charger is in top-off mode, CHG_OK = 1, V _{MBATT} ≥ V _{BATREG} , T _J < T _{JSHDN} 0x04 = Charger is in done mode, CHG_OK=0, V _{MBATT} > V _{BATREG} - V _{RSTRT} , T _J < T _{JSHDN} 0x06 = Charger is in timer fault mode, CHG_OK = 0, V _{MBATT} < V _{BATOV} , if BAT_DTLS = 0b001 then V _{MBATT} < V _{BATPQ} , T _J < T _{JSHDN} 0x07 = Charger is in DETBAT = High suspend mode, CHG_OK = 0, V _{MBATT} < V _{BATOV} , if BAT_DTLS = 0b001 then V _{MBATT} < V _{PQLB} , T _J < T _{JSHDN} 0x08 = Charger is off, charger input invalid and/or charger is disabled, CHG_OK = 1. 0x09 = Reserved 0x0A = Charger is off and the junction temperature is > T _{JSHDN} , CHG_OK = 0. 0x0B = Charger is off because the watchdog timer expired, CHG_OK = 0.
6:4	R	BAT_DTLS	000	Battery Details 0x00 = No battery and the charger is suspended. 0x01 = V _{MBATT} < V _{PQLB} . This condition is also reported in the CHG_DTLS as 0x00 0x02 = The battery is taking longer than expected to charge. This could be due to high system currents, an old battery, a damaged battery or something else. Charging has suspended and the charger is in its timer fault mode. This condition is also reported in the CHG_DTLS as 0x06. 0x03 = The battery is okay and its voltage is greater than the minimum system voltage (V _{SYSMIN} < V _{MBATT}), QBAT is on and V _{SYS} is approximately equal to V _{MBATT} . 0x04 = The battery is okay but its voltage is low: V _{PQLB} < V _{MBATT} < V _{SYSMIN} . Q _{BAT} is operating like an LDO to regulate V _{SYS} to V _{SYSMIN} . 0x05 = The battery voltage is greater than the battery overvoltage flag threshold (V _{BATOVF}) or it has been greater than this threshold within the last 37.5ms. V _{BATOVF} is set to a percentage above the V _{BATREG} target as programmed by CHG_CV_PRM. Note that this flag is only be generated when there is a valid input or when the DC-DC is operating as a boost. 0x06 = The battery is overcurrent or it has been overcurrent for at least 6ms since the last time this register has been read. 0x07 = Reserved In the event that multiple faults occur within the battery details category, overcurrent has priority followed by no battery, then overvoltage, then timer fault, then below prequel.

CHG_DETAILS_01 Register Bit Description (0xB4) (continued)

BIT	MODE	NAME	RESET	DESCRIPTION
7	R	TREG	0	Temperature Regulation Status 0 = The junction temperature is less than the threshold set by REGTEMP and the full charge current limit is available. 1 = The junction temperature is greater than the threshold set by REGTEMP and the charge current limit may be folding back to reduce power dissipation.

CHG_DETAILS_02 Register Bit Description (0xB5)

NAME	FUNCTION	ADDR	ТҮРЕ	RESET
CHG_DTLS_02	Charger details 02	0xB5	0	0x00

BIT	MODE	NAME	RESET	DESCRIPTION
3:0	R	BYP_DTLS	0000	Bypass Node Details. All bits in this family are independent from each other. They are grouped together only because they all relate to the health of the BYP node and any change in these bits generates a BYP_I interrupt. BYP_DTLS0 = OTGILIM = 0bxxx1 BYP_DTLS1 = BSTILIM = 0bxx1x BYP_DTLS2 = BCKNegILIM = 0bx1xx *********************************
7:4	R	RSVD	00	Reserved

CHG_CNFG_00 Register Bit Description (0xB7)

NAME	FUNCTION	ADDR	TYPE	RESET
CHG_CNFG_00	Charger configuration 00	0xB7	0	0x05

BIT	MODE	NAME	RESET	DESCRIPTION
3:0	R/W	MODE	0101	Smart Power Selector Configuration 0x00 = 0b00000 = charger = off, OTG = off, buck = off, boost = off. The QBAT switch is on to allow the battery to support the system. BYP may or may not be biased based on the CHGIN availability. 0x01 = 0b0001 = same as 0b0000 0x02 = 0b0010 = same as 0b0000 0x04 = 0b0100 = charger = off, OTG = off, buck = on, boost = off. When there is a valid input, the buck converter regulates the system voltage to be VBATREG. 0x05 = 0b0101 = charger = on, OTG = off, buck = on, boost = off. When there is a valid input, the battery is charging. VSYS is the larger of VSYSMIN and ~VMBATT + IMBATT x RBAT2SYS. 0x06 = 0b0110 = same as 0b101 0x07 = 0b0111 = same as 0b101 0x08 = 0b1000 = charger = off, OTG = off, buck = off, boost = on. The QBAT switch is on to allow the battery to support the system and the charger's DC-DC operates as a boost converter. The BYP voltage is regulated to VBYPSET. QCHGIN is off. 0x09 = 0b1010 = charger = off, OTG = on, buck = off, boost = on. The QBAT switch is on to allow the battery to support the system, the charger's DC-DC operates as a boost converter. QCHGIN is on allowing it to source current up to ICHGIN.OTG.MAX. The boost target voltage is 5.1V (VBYPOTG). 0x0B = reserved 0x0C = 0b1100 = charger = off, OTG = off, buck = on, boost = on. When there is a valid input, the system is supported from that input: VSYS = 4.2V. When input is invalid, the boost is on with a target voltage that is VBYPSET. 0x0D = 0b1101 = charger = on, OTG = off, buck = on, boost = on. When there is a valid input, the system is supported from that input: VSYS is the larger of VSYSMIN and ~VMBATT + IMBATT x RBAT2SYS. When input is invalid, the boost is on with a target voltage that input: VSYS is the larger of VSYSMIN and ~VMBATT + IMBATT x RBAT2SYS. When input is invalid, the boost is on with a target voltage of 5.1V (VBYP.OTG). 0x0F = 0b1111 = charger = on, OTG = on, buck = on, boost = on. When there is a valid WCIN input, the system is supported from that input: VSYS is the

CHG_CNFG_00 Register Bit Description (0xB7) (continued)

BIT	MODE	NAME	RESET	DESCRIPTION
4	R/W	WDTEN	0	Watchdog Timer Enable Bit. While enabled, the system controller must reset the watchdog timer within the timer period (t _{WD}) for the charger to operate normally. Reset the watchdog timer by programming WDTCLR = 0x01. 0 = Watchdog timer disabled 1 = Watchdog timer enabled
5	R/W	SPREAD	0	Spread Spectrum Feature Feature is operational both for 9V and 12V CHGIN input voltage. Feature is not guaranteed to be operational for 5V CHGIN/WCIN input voltage. When feature is not operational, it can be kept enabled without side effect. 0: Disabled 1: Enabled
6	R/W	DISIBS	0	MBATT to SYS FET Disable Control 0 = MBATT to SYS FET is controlled by the power path state machine. 1 = MBATT to SYS FET is forced off.
7	R/W	OTG_ CTRL	0	OTG FET Control 0 = Forces the CHGIN input switch to be on when in Mode = 0x0E or 0x0F. 1 = The CHGIN turns off anytime power switches between WCIN and BATT.

CHG_CNFG_01 Register Bit Description (0xB8)

NAME	FUNCTION ADDR		TYPE	RESET
CHG_CNFG_01	Charger details 01	0xB8	O R/W (protected with CHGPROT)	0x10

BIT	MODE	NAME	RESET	DESCRIPTION			
2:0	R/W	FCHGTIME	001	Fast-Charge Timer Duration (t _{FC}) 0x00 = disable 0x01 = 4hrs 0x02 = 6hrs 0x03 = 8hrs 0x04 = 10hrs 0x05 = 12hrs 0x06 = 14hrs 0x07 = 16hrs			
3	R/W	FSW	Switching Frequency Option 0 0: 4MHz 1: 2MHz				
5:4	R/W	CHG_RSTRT	01	Charger Restart Threshold 0x00 = 100mV below the value programmed by CHG_CV_PRM 0x01 = 150mV below the value programmed by CHG_CV_PRM 0x02 = 200mV below the value programmed by CHG_CV_PRM 0x03 = Disabled			
6	R/W	LSEL	0	Inductor Selection 0:0.47µH (for 4MHz without restriction) 2MHz/0.47µH option can be used only in the charge mode and is forbidden in following use cases: Boost mode Buck mode 1:1µH (for 2MHz and 4MHz option)			
7	R/W	PQEN	0	Low Battery Prequalification Mode Enable 0 = Low battery prequalification mode is disabled. 1 = Low battery prequalification mode is enabled.			

CHG_CNFG_02 Register Bit Description (0xB9)

NAME	FUNCTION	ADDR	ТҮРЕ	RESET		
CHC CNEC 02	Charger	000	Charger O		0x09	
CHG_CNFG_02	configuration 02	0xB9	R/W (protected with CHGPROT)			

BIT	MODE	NAME	RESET	DESCR	DESCRIPTION						
				Fast-Charge Current Selection. When the charger is enabled, the charge current limit is set by these bits. These bits range from 0.10A (0x00) to 3.0A (0x3C) in 50mA step. Note that the first 3 codes are all 100mA							
				Bits	(mA)	Bits	(mA)	Bits	(mA)	Bits	(mA)
				0x00	100	0x10	800	0x20	1600	0x30	2400
				0x01	100	0x11	850	0x21	1650	0x31	2450
				0x02	100	0x12	900	0x22	1700	0x32	2500
				0x03	150	0x13	950	0x23	1750	0x33	2550
				0x04	200	0x14	1000	0x24	1800	0x34	2600
				0x05	250	0x15	1050	0x25	1850	0x35	2650
	5.044		001001	0x06	300	0x16	1100	0x26	1900	0x36	2700
5:0	R/W	CHG_CC	(450mA)	0x07	350	0x17	1150	0x27	1950	0x37	2750
				0x08	400	0x18	1200	0x28	2000	0x38	2800
				0x09	450	0x19	1250	0x29	2050	0x39	2850
				0x0A	500	0x1A	1300	0x2A	2100	0x3A	2900
				0x0B	550	0x1B	1350	0x2B	2150	0x3B	2950
				0x0C	600	0x1C	1400	0x2C	2200	0x3C	3000
				0x0D	650	0x1D	1450	0x2D	2250	0x3D	3000
				0x0E	700	0x1E	1500	0x2E	2300	0x3E	3000
				0x0F	750	0x1F	1550	0x2F	2350	0x3F	3000
					it the ther		ack loop ca	an reduce	the batte	ry chargei	r's target
7:6	R/W	OTG_ILIM	00	CHGIN Output Current Limit in OTG Mode (I _{CHGIN.OTG.LIM}) When MODE = 0x09 or 0x0A the CHGIN current limit is set at the following current limit: 00 = 500mA 01 = 900mA 10 = 1200mA 11 = 1500mA							

CHG_CNFG_03 Register Bit Description (0xBA):

NAME	FUNCTION	ADDR	TYPE	RESET
CHG_CNFG_03	Charger configuration 03	0xBA	O R/W (protected with CHGPROT)	0xDA

BIT	MODE	NAME	RESET	DESCRIPTION
2:0	R/W	TO_ITH	010 (150mA)	Top-Off Current Threshold. The charger transitions from its fast-charge constant voltage mode to its top-off mode when the charger current decays to the value programmed by this register. This transition generates a CHG_I interrupt and causes the CHG_DTLS register to report top-off mode. This transition also starts the top-off time as programmed by TO_TIME. 0x00 = 0.1A 0x01 = 0.125A 0x02 = 0.15A 0x03 = 0.175A 0x04 = 0.2A 0x05 = 0.25A 0x06 = 0.3A 0x07 = 0.35A
5:3	R/W	TO_TIME	011 (30min)	Top-Off Timer Setting 0x00 = 0min 0x01 = 10min 0x02 = 20min 0x03 = 30min 0x04 = 40min 0x05 = 50min 0x06 = 60min 0x07 = 70min
7:6	R/W	ILIM	11	Program Buck Peak Current Limit 00: support ICHG=3.00A 01: support ICHG=2.75A 10: support ICHG= 2.50A 11: support ICHG=2.25A

CHG_CNFG_04 Register Bit Description (0xBB):

NAME	FUNCTION	ADDR	ТҮРЕ	RESET
CHG_CNFG_04	Charger configuration 04	0xBB	O R/W (protected with CHGPROT)	0x96

BIT	MODE	NAME	RESET	DESCRIPT	ION					
				Primary Charge Termination Voltage Setting When the charger is enabled and the main-battery temperature is < T3 if JEITA = 1 or < T4 if JEITA = 0, then, the charger's battery regulation voltage (VBATREG) is set by CHG_CV_PRM.						
				BITS	V	BITS	V	BITS	V	
				0x00	3.650	0x10	4.050	0x20	4.425	
				0x01	3.675	0x11	4.075	0x21	4.450	
				0x02	3.700	0x12	4.100	0x22	4.475	
	R/W			0x03	3.725	0x13	4.125	0x23	4.500	
		CHG_CV_PRM		0x04	3.750	0x14	4.150	0x24	4.525	
5:0			010110 (4.2V)	0x05	3.775	0x15	4.175	0x25	4.550	
				0x06	3.800	0x16	4.200	0x26	4.575	
				0x07	3.825	0x17	4.225	0x27	4.600	
				0x08	3.850	0x18	4.250	0x28	4.625	
				0x09	3.875	0x19	4.275	0x29	4.650	
				0x0A	3.900	0x1A	4.300	0x2A	4.675	
				0x0B	3.925	0x1B	4.325	0x2B	4.700	
				0x0C	3.950	0x1C	4.340			
				0x0D	3.975	0x1D	4.350			
				0x0E	4.000	0x1E	4.375			
				0x0F	4.025	0x1F	4.400			
7:6	R/W	MINVSYS	10 (3.6V)	Minimum System Regulation Voltage (V _{SYSMIN}) $0x00 = 3.4V$ $0x01 = 3.5V$ $0x02 = 3.6V$ $0x03 = 3.7V$						

CHG_CNFG_06 Register Bit Description (0xBD):

NAME	FUNCTION	ADDR	TYPE	RESET
CHG_CNFG_06	Charger configuration 06	0xBD	0	0x00

BIT	MODE	NAME	RESET	DESCRIPTION
1:0	R/W	WDTCLR	00	Watchdog Timer Clear Bits Writing 01 to these bits clears the watchdog timer when the watchdog timer is enabled. 0x00 = the watchdog timer is not cleared 0x01 = the watchdog timer is cleared 0x02 = the watchdog timer is not cleared 0x03 = the watchdog timer is not cleared
3:2	R/W	CHGPROT	00	Charger Settings Protection Bits Writing 11 to these bits unlocks the write capability for the registers who are protected with CHGPROT writing any value besides 11 locks these registers. 0x00 = write capability is locked 0x01 = write capability is locked 0x02 = write capability is locked 0x03 = write capability is unlocked
7:4	R/W	RSVD	0000	Reserved

CHG_CNFG_07 Register Bit Description (0xBE):

NAME	FUNCTION	ADDR	TYPE	RESET
CHG_CNFG_	7 Charger configuration 07	0xBE	O R/W (protected with CHGPROT)	0x40

BIT	MODE	NAME	RESET	DESCRIPTION
4:0	R/W	RSVD	00000	Reserved
6:5	R/W	REGTEMP	10	Junction Temperature Thermal Regulation Loop Set point. The charger's target current limit starts to foldback and the TREG bit is set if the junction temperature is greater than the REGTEMP setpoint. 0x00 = 85°C 0x01 = 100°C 0x02 = 115°C 0x03 = 130°C
7	R/W	RSVD	0	Reserved

CHG_CNFG_09 Register Bit Description (0xC0):

NAME	FUNCTION	ADDR	TYPE	RESET
CHG_CNFG_09	Charger configuration 09	0xC0	0	0x0F

BIT	MODE	NAME	RESET	DESCR	IPTION							
				Maximum Input Current Limit Selection 7-bit adjustment from 100mA to 4.0A in 33mA steps. Note that the first 4 codes are all 100mA:								
				Bits	Unit (mA)	Bits	Unit (mA)	Bits	Unit (mA)	Bits	Unit (mA)	
				0x00	100	0x20	1067	0x40	2133	0x60	3200	
				0x01	100	0x21	1100	0x41	2167	0x61	3233	
				0x02	100	0x22	1133	0x42	2200	0x62	3267	
				0x03	100	0x23	1167	0x43	2233	0x63	3300	
				0x04	133	0x24	1200	0x44	2267	0x64	3333	
				0x05	167	0x25	1233	0x45	2300	0x65	3367	
				0x06	200	0x26	1267	0x46	2333	0x66	3400	
				0x07	233	0x27	1300	0x47	2367	0x67	3433	
	R/W			0x08	267	0x28	1333	0x48	2400	0x68	3467	
				0x09	300	0x29	1367	0x49	2433	0x69	3500	
				0x0A	333	0x2A	1400	0x4A	2467	0x6A	3533	
		CHGIN_ILIM		0x0B	367	0x2B	1433	0x4B	2500	0x6B	3567	
				0x0C	400	0x2C	1467	0x4C	2533	0x6C	3600	
6:0			0x0F	0x0D	433	0x2D	1500	0x4D	2567	0x6D	3633	
			(0.50A)	0x0E	467	0x2E	1533	0x4E	2600	0x6E	3667	
				0x0F	500	0x2F	1567	0x4F	2633	0x6F	3700	
				0x10	533	0x30	1600	0x50	2667	0x70	3733	
				0x11	567	0x31	1633	0x51	2700	0x71	3767	
				0x12	600	0x32	1667	0x52	2733	0x72	3800	
				0x13	633	0x33	1700	0x53	2767	0x73	3833	
				0x14	667	0x34	1733	0x54	2800	0x74	3867	
				0x15	700	0x35	1767	0x55	2833	0x75	3900	
				0x16	733	0x36	1800	0x56	2867	0x76	3933	
				0x17	767	0x37	1833	0x57	2900	0x77	3967	
				0x18	800	0x38	1867	0x58	2933	0x78	4000	
				0x19	833	0x39	1900	0x59	2967	0x79	4000	
				0x1A	867	0x3A	1933	0x5A	3000	0x7A	4000	
				0x1B	900	0x3B	1967	0x5B	3033	0x7B	4000	
				0x1C	933	0x3C	2000	0x5C	3067	0x7C	4000	
				0x1D	967	0x3D	2033	0x5D	3100	0x7D	4000	
				0x1E	1000	0x3E	2067	0x5E	3133	0x7E	4000	
				0x1F	1033	0x3F	2100	0x5F	3167	0x7F	4000	
7	R/W	RSVD	0	Reserve	d							

CHG_CNFG_10 Register Bit Description (0xC1):

NAME	FUNCTION	ADDR	TYPE	RESET
CHG_CNFG_10	Charger configuration 09	0xC1	0	0x19

BIT	MODE	NAME	RESET	DESCRIPTION			
				Maximum Input Current Limit Selection 6bit adjustment from 60mA to 1.260A in 20mA steps. Note that the first 4 codes are all 60mA:			
				Bits	(mA)	Bits	(mA)
				0x00	60	0x20	640
				0x01	60	0x21	660
				0x02	60	0x22	680
				0x03	60	0x23	700
				0x04	80	0x24	720
				0x05	100	0x25	740
				0x06	120	0x26	760
				0x07	140	0x27	780
				0x08	160	0x28	800
				0x09	180	0x29	820
				0x0A	200	0x2A	840
				0x0B	220	0x2B	860
	5:0 R/W V			0x0C	240	0x2C	880
		WCIN_ILIM	0x19	0x0D	260	0x2D	900
5:0			(0.50A)	0x0E	280	0x2E	920
				0x0F	300	0x2F	940
				0x10	320	0x30	960
				0x11	340	0x31	980
				0x12	360	0x32	1000
				0x13	380	0x33	1020
				0x14	400	0x34	1040
				0x15	420	0x35	1060
				0x16	440	0x36	1080
				0x17	460	0x37	1100
				0x18	480	0x38	1120
				0x19	500	0x39	1140
				0x1A	520	0x3A	1160
				0x1B	540	0x3B	1180
				0x1C	560	0x3C	1200
				0x1D	580	0x3D	1220
				0x1E	600	0x3E	1240
				0x1F	620	0x3F	1260
7:6	R/W	RSVD	00	Reserved			

CHG_CNFG_11 Register Bit Description (0xC2):

NAME	FUNCTION	ADDR	TYPE	RESET
CHG_CNFG_11	Charger configuration 11	0xC2	0	0x00

BIT	MODE	NAME	RESET	DESCR	IPTION						
				3V (0x0	Target Out 0) to 5.8V ode (MODI	(0x70) in	0.025V ste		setting is v	alid for the	e "boost
				Bits	Unit (mA)	Bits	Unit (mA)	Bits	Unit (mA)	Bits	Unit (mA)
				0x00	3.000	0x20	3.800	0x40	4.600	0x60	5.400
				0x01	3.025	0x21	3.825	0x41	4.625	0x61	5.425
				0x02	3.050	0x22	3.850	0x42	4.650	0x62	5.450
				0x03	3.075	0x23	3.875	0x43	4.675	0x63	5.475
				0x04	3.100	0x24	3.900	0x44	4.700	0x64	5.500
				0x05	3.125	0x25	3.925	0x45	4.725	0x65	5.525
				0x06	3.150	0x26	3.950	0x46	4.750	0x66	5.550
				0x07	3.175	0x27	3.975	0x47	4.775	0x67	5.575
				0x08	3.200	0x28	4.000	0x48	4.800	0x68	5.600
				0x09	3.225	0x29	4.025	0x49	4.825	0x69	5.625
				0x0A	3.250	0x2A	4.050	0x4A	4.850	0x6A	5.650
		VBYPSET		0x0B	3.275	0x2B	4.075	0x4B	4.875	0x6B	5.675
			0.00	0x0C	3.300	0x2C	4.100	0x4C	4.900	0x6C	5.700
6:0	R/W		0x00 (3V)	0x0D	3.325	0x2D	4.125	0x4D	4.925	0x6D	5.725
			(30)	0x0E	3.350	0x2E	4.150	0x4E	4.950	0x6E	5.750
				0x0F	3.375	0x2F	4.175	0x4F	4.975		
				0x10	3.400	0x30	4.200	0x50	5.000		
				0x11	3.425	0x31	4.225	0x51	5.025		
				0x12	3.450	0x32	4.250	0x52	5.050		
				0x13	3.475	0x33	4.275	0x53	5.075		
				0x14	3.500	0x34	4.300	0x54	5.100		
				0x15	3.525	0x35	4.325	0x55	5.125		
				0x16	3.550	0x36	4.350	0x56	5.150		
				0x17	3.575	0x37	4.375	0x57	5.175		
				0x18	3.600	0x38	4.400	0x58	5.200		
				0x19	3.625	0x39	4.425	0x59	5.225		
				0x1A	3.650	0x3A	4.450	0x5A	5.250		
				0x1B	3.675	0x3B	4.475	0x5B	5.275		
				0x1C	3.700	0x3C	4.500	0x5C	5.300		
				0x1D	3.725	0x3D	4.525	0x5D	5.325		
				0x1E	3.750	0x3E	4.550	0x5E	5.350		
				0x1F	3.775	0x3F	4.575	0x5F	5.375		
7	R/W	RSVD	0	Reserve	d						

CHG_CNFG_12 Register Bit Description (0xC3):

NAME	FUNCTION	ADDR	TYPE	RESET
CHG_CNFG_12	Charger configuration 12	0xC3	0	0x67

BIT	MODE	NAME	RESET	DESCRIPTION
2:0	R/W	B2SOVRC	111	BAT to SYS Overcurrent Threshold 0x00 = Disabled 0x01 = 3.00A 0x02 = 3.25A 0x03 = 3.50A 0x04 = 3.75A 0x05 = 4.00A 0x06 = 4.25A 0x07 = 4.50A
4:3	R/W	VCHGIN_REG	00	CHGIN Voltage Regulation Threshold (V _{CHGIN_REG}) Adjustment The CHGIN to GND Minimum Turn-On Threshold (V _{CHGIN_UVLO}) also scales with this adjustment. 0x00 = V _{CHGIN_REG} = 4.3V and V _{CHGIN_UVLO} = 4.5V 0x01 = V _{CHGIN_REG} = 4.7V and V _{CHGIN_UVLO} = 4.9V 0x02 = V _{CHGIN_REG} = 4.8V and V _{CHGIN_UVLO} = 5.0V 0x03 = V _{CHGIN_REG} = 4.9V and V _{CHGIN_UVLO} = 5.1V
5	R/W	CHGINSEL	1	CHGIN/USB Input Channel Select 0 = Disabled 1 = Enabled
6	R/W	WCINSEL	1	WCIN Input Channel Select 0 = Disabled 1 = Enabled
7	R/W	Reserved	0	Reserved

Model Gauge m5 Register Map

TAIrtTh2 (0xB2)

BITFIELD	BITS	DESCRIPTION	DECODE
TempWarm	15:8	Temperature threshold used for smart charging as T4	Units of LSB are 1°C. Set to 0x7F to disable.
TempCool	7:0	Temperature threshold used for smart charging as T1	Units of LSB are 1°C. Set to 0x80 to disable.

SmartChgCfg (0xDB)

BITFIELD	BITS	DESCRIPTION	DECODE
DisJEITA	5	Set 1 to disable JEITA battery temperature monitor adjusts	
UseVF	4		0x0: MixSOC is an input SOC for SmartCharging 0x1: VFSOC is an input SOC for SmartCharging
EnsC	1	Set 1 to enable SmartCharing	
EnSF	0	Set 1 to enable SmartFull	

Status and Configuration Registers

Register Details

Status (0x00)

Interrupt status register for the FG block.

BITFIELD	BITS	DESCRIPTION	DECODE
Br	15	Battery Removal	This bit is set to 1 when the device detects that a battery has been removed from the system. This bit must be cleared by system software to detect the next removal event. Br is set to 0 at power-up.
Smx	14	Maximum SOCALRT Threshold Exceeded	This bit is set to 1 whenever SOC rises above the maximum SOCALRT value. This bit may or may not need to be cleared by system software to detect the next event. See SS in the CONFIG register and SACFG in the MiscCFG register. Smx is set to 0 at power-up.
Tmx	13	Maximum TALRT Threshold Exceeded	This bit is set to 1 whenever a Temperature register reading is above the maximum TALRT value. This bit may or may not need to be cleared by system software to detect the next event. See TS in the CONFIG register. Tmx is set to 0 at power-up.
Vmx	12	Maximum VALRT Threshold Exceeded	This bit is set to 1 whenever a VCELL register reading is above the maximum VALRT value. This bit may or may not need to be cleared by system software to detect the next event. See VS in the CONFIG register. Vmx is set to 0 at power-up.
Bi	11	Battery Insertion	This bit is set to 1 when the device detects that a battery has been inserted into the system by monitoring the AIN pin. This bit must be cleared by system software to detect the next insertion event. Bi is set to 0 at power-up.

BITFIELD	BITS	DESCRIPTION	DECODE
Smn	10	Minimum SOCALRT Threshold Exceeded	This bit is set to 1 whenever SOC falls below the minimum SOCALRT value. This bit may or may not need to be cleared by system software to detect the next event. See SS in the CONFIG register and SACFG in the MiscCFG register. Smn is set to 0 at power-up.
Tmn	9	Minimum TALRT Threshold Exceeded	This bit is set to 1 whenever a Temperature register reading is below the minimum TALRT value. This bit may or may not need to be cleared by system software to detect the next event. See TS in the CONFIG register. Tmn is set to 0 at power-up.
Vmn	8	Minimum VALRT Threshold Exceeded	This bit is set to 1 whenever a VCELL register reading is below the minimum VALRT value. This bit may or may not need to be cleared by system software to detect the next event. See VS in the CONFIG register. Vmn is set to 0 at power-up.
dSOCi	7	1% SOC Change Alert	This bit is set to 1 to indicate a 1% SOC change alert. dSOCi is set to 0 at power-up.
ThmHot	6	FG Control Charger Input Current Limit	Set to 1 to indicate a Thermistor Hot to allow FG control charger input current limit. ThmHot is set to 0 at power-up.
SPR_5	5		
Isysmx	4	SYS current is over OCP limit.	Maximum SYS current threshold exceeded.
Bst	3	Battery Status	This bit is set to 0 when a battery is present in the system and set to 1 when the battery is removed. Bst is set to 0 at power-up.
SPR_2	2		
POR	1	Power-On Reset	This bit is set to 1 when the device detects that a software or hardware POR event has occurred. If the host detects that the POR bit has been set, the device should be reconfigured. See the Power-Up and Power-Up and Power-Up and Power-Up and <a href<="" td="">
Imn	0	Minimum Isys Threshold Exceeded	

VAIrtTh (0x01)

BITFIELD	BITS	DESCRIPTION	DECODE
MaxVoltageAlrt	15:8	Sets an alert threshold for maximum voltage.	Set Max = 0xFF to disable. Units of LSB are 20mV.
MinVoltageAlrt	7:0	Sets an alert threshold for minimum voltage.	Set Min = 0x00 to disable. Units of LSB are 20mV.

TAIrtTh (0x02)

BITFIELD	BITS	DESCRIPTION	DECODE
MaxTempAlrt	15:8	Sets an alert threshold for maximum temperature.	Set Max = 0x7F to disable. Units of LSB are 1°C.
MinTempAlrt	7:0	Sets an alert threshold for minimum temperature.	Set Min = 0x80 to disable. Units of LSB are 1°C.

SAIrtTh (0x03)

BITFIELD	BITS	DESCRIPTION	DECODE
MaxSocAlrt	15:8	Sets an alert for maximum SOC.	This may be used for charge/discharge termination, or for power-management near empty. Set Max = 0xFF to disable. Units of LSB are 1%.
MinSocAlrt	7:0	Sets an alert for minimum SOC.	This may be used for charge/discharge termination, or for power-management near empty. Set Min = 0x00 to disable. Units of LSB are 1%.

AtRate (0x04)

BITFIELD	BITS	DESCRIPTION
AtRate	15:0	Host software should write the AtRate register with a negative two's-complement 16-bit value of a theoretical load current prior to reading any of the at-rate output registers (AtTTE, AtAvSOC, AtAvCap).

QRTable00 (0x12)

BITFIELD	BITS	DESCRIPTION
QRTable00	15:0	QRTable is interpreted the same as BT07, using interpolation. However, during empty learning 2 QRTable registers should be modified with interpolation applied.

FullSocThr (0x13)

BITFIELD	BITS	DESCRIPTION	DECODE
FullSOCThr	15:0	FullSOCThr comes from OTP if the OTP register is enabled. Otherwise, it POR's to a default of 95%.	LSB unit is 1/256%.

DesignCap (0x18)

BITFIELD	BITS	DESCRIPTION	DECODE
DesignCap	15:0	DesignCap16 is used to relate to FullCap16 to help measure the age of the battery.	Units are 0.5mAh with 10mΩ sense resistor.

Config (0x1D)

BITFIELD	BITS	DESCRIPTION	DECODE
SPR_15	15	Fuel-Gauge Charger Fail Enable	Set to 1 to enable FGCHGFAIL interrupt to drive the INTB pin. This bit is not accessible by the fuel-gauge firmware.
Ss	14	SOC ALRT Sticky	When SS = 1, SOC alerts can only be cleared through software. When SS = 0, SOC alerts are cleared automatically when the threshold is no longer exceeded. SS is set to 0 at power-up.
Ts	13	Temperature ALRT Sticky	When TS = 1, temperature alerts can only be cleared through software. When TS = 0, temperature alerts are cleared automatically when the threshold is no longer exceeded. TS is set to 1 at power-up.
Vs	12	Voltage ALRT Sticky	When VS = 1, voltage alerts can only be cleared through software. When VS = 0, voltage alerts are cleared automatically when the threshold is no longer exceeded. VS is set to 0 at power-up.
SPR_11	11	Fuel-Gauge Charger Control	Set to 1 to enable MaxCharge (fuel-gauge controlled charging).
AINSH	10	AIN Pin Shutdown	Set to 1 to enable device shutdown when the battery is removed. The IC enters shutdown if the AIN pin remains high (AIN reading > VTHRM - VDETR) for longer than the timeout of the SHDN-TIMER register. This also configures the device to wake up when AIN is pulled low on cell insertion. AINSH is set to 0 at power-up. Note that if I2CSH and AINSH are both set to 0, the device wakes up an edge of any of the SDA, SCL, or INTB pins.
Ten	9	Enable Temperature Channel	Set to 1 and set ETHRM or FTHRM to 1 to enable measurements on the AIN pin. Ten is set to 1 at power-up.
Tex	8	Temperature External	When set to 1, the fuel gauge requires external temperature measurements to be written from the host. When set to 0, measurements on the AIN pin are converted to a temperature value and stored in the Temperature register instead. Tex is set to 1 at power-up.
SHDN	7	Shutdown	Write this bit to logic 1 to force a shutdown of the device after timeout of the SHDNTIMER register. SHDN is reset to 0 at power-up and upon exiting shutdown mode

BITFIELD	BITS	DESCRIPTION	DECODE
I2CSH	6	I ² C Shutdown	Set to logic 1 to force the device to enter shutdown mode if both SDA and SCL are held low for more than timeout of the SHDNTIMER register. This also configures the device to wake up on a rising edge of either SDA or SCL. Set to 1 at power-up. Note that if I2CSH and AINSH are both set to 0, the device wakes up an edge of any of the SDA, SCL, or INTB pins.
SPR_5	5	I ² C Charge Fail Enable	Set to 1 to enable I2CChgFail interrupt to drive the INTB pin. This bit is not accessible by the fuel-gauge firmware.
ETHRM	4	Enable Thermistor	Set to logic 1 to enable the automatic THRM output bias and AIN measurement every 1.4s. This bit is set to 1 at power-up.
FTHRM	3	Force Thermistor Bias Switch. This allows the host to control the bias of the thermistor switch or enable fast detection of battery removal (see the <u>Fast Detection of Cell Removal</u> section).	Set FTHRM = 1 to always enable the thermistor bias switch. With a standard $10k\Omega$ thermistor, this adds an additional ~200 μ A to the current drain of the circuit. This bit is set to 0 at power-up.
Aen	2	Enable alert on fuel-gauge outputs.	When Aen = 1, violation of any of the alert threshold register values by temperature, voltage, or SOC triggers an alert. This bit affects the INTB pin (FG_INT) operation only. The Smx, Smn, Tmx, Tmn, Vmx, and Vmn bits are not disabled. This bit is set to 0 at power-up. Note that if this bit is set to 1, the ALSH bit should be set to 0 to prevent an alert condition from causing the device to enter shutdown mode.
Bei	1	Enable alert on battery insertion.	When Bei = 1, a battery-insertion condition, as detected by the AIN pin voltage, triggers an alert. Set to 0 at power-up. Note that if this bit is set to 1, the ALSH bit should be set to 0 to prevent an alert condition from causing the device to enter shutdown mode.
Ber	0	Enable alert on battery removal.	When Ber = 1, a battery-removal condition, as detected by the AIN pin voltage, triggers an alert. Set to 0 at power-up. Note that if this bit is set to 1, the ALSH bit should be set to 0 to prevent an alert condition from causing the device to enter shutdown mode.

IChgTerm (0x1E)

BITFIELD	BITS	DESCRIPTION	DECODE
ICHGTerm	15:0	No change from BT07. ICHGTerm restores from OTP if enabled, however it restores according to nDesignCap/10 if OTP is disabled for ICHGTerm.	LSB unit is the same as the register current's LSB unit.

DevName (0x21)

BITFIELD	BITS	DESCRIPTION	
DevName	15:0	Firmware version information	

QRTable10 (0x22)

BITFIELD	BITS	DESCRIPTION
QRTable10	15:0	QRTable is interpreted the same as BT07, using interpolation. However, during empty learning 2 QRTable registers should be modified with interpolation applied.

FullCapNom (0x23)

BITFIELD	BITS	DESCRIPTION	DECODE
FullCapNom	15:0	FullCapNom is the internally measured value of the nominal full capacity estimated for room temperature. It is measured by one of the 3 defined full capacity learning methods (relaxto-relax, relax-to-relax zigzag, or continual).	$0.5 mAh$ per LSB with $10 m\Omega$ sense resistor.

TempNom (0x24)

BITFIELD	BITS	DESCRIPTION	DECODE
TempNom	15:6	When Temperature measurements are hotter than TempNom, the VFG is compensated using TempCoHot. When Temperature measurements are colder than TempNom, the VFG is compensated using TempCoCold.	10 bit value in bits D15 through D6. The MSbit in D15 represents sign which is useful for values in degrees C but should always be set to 0 for ratiometric values. The LSB is 0.0039°C and the upper Byte has units 1°C. This is a signed register.
SPR	5:0		

TempLim (0x25)

BITFIELD	BITS	DESCRIPTION	DECODE
TempHot	15:8	RCOMP0 is learned when TempCold <avgta<temphot. (tempnom+temphot)="" 2<avgt="" a<temphot.<="" is="" learned="" td="" tempcohot="" when=""><td>Units are 1°C.</td></avgta<temphot.>	Units are 1°C.
TempCold	7:0	RCOMP0 is learned when TempCold <avgta<temphot.tempcocold 1°c.<="" 2.="" are="" is="" learned="" td="" tempcold<avgta,(tempnom+tempcold)="" units="" when=""><td>Units are 1°C.</td></avgta<temphot.tempcocold>	Units are 1°C.

LearnCfg (0x28)

The LearnCFG register controls all functions relating to adaptation during operation. The LearnCFG register default values should not be changed unless specifically required by the application.

BITFIELD	BITS	DESCRIPTION	DECODE
Learn- RCOMP	15:13	LearnRCOMP defines the rate to learn RCOMP0.	
LearnTCO	12:10	LearnTCO defines the rate to learn the temperature compensation. Set LearnRCOMP = LearnTCO = 0 to disable learning.	Set to 1 for increment/decrement learning. Otherwise, any non-zero value applies the error signal (VFOCVprerelax – VFOCVpostrelax) to the learn parameter (RCOMP0, TempCoUp, or TempCoDn) by left-shifting the learn level (LearnRCOMP or LearnTCO).
FCLm	9:8	FCLrnMethod defines one of 3 methods for learning FullCAP:0) relax-to-relax (DS2786 style). d%_acc16 and dQacc16 are only calculated at relaxation points, and FullCAP is only calculated at these relax points. However, unlike the DS2786, if the delta-percentage is too small to successfully learn capacity, the measurement is accumulated towards the next relax. Set FCL-rnMethod = 1 to enable relax-to-relax learning. This method carries the least error, although it is the least opportunistic. 1) Not applicable. 2) Zigzag. d%_acc16 and dQacc16 are calculated frequently. FullCAP learning only occurs during relax. This improves accuracy of d%_acc16 and reduces error contributed by voltage fuel-gauge. Set FCLrnMethod = 2 to enable ZigZag learning. 3) Continual learning. d%_acc16 and dQacc16 are calculated frequently. FullCAP occurs during any d%_acc16 overflow. Relax criteria not needed. Set FCLrnMethod = 3 to enable continual learning.	FCLrn0 is FCalways. Set FCalways = 1 to enable fullcap learning even when relax is not detected. FCLrn1 is accalways. Set accalways = 1 to enable dQ and dacc accumulation even when relax is not detected. Set accalways = 0 to enable dQ/dAcc accumulation only when relax is detected.
FCx	7	FCx sets the source value to effect FullCAP when charge termination is detected.	Set FCx = 0 (follow RepCap)

BITFIELD	BITS	DESCRIPTION				DECODE
FCLrnStage	BITS 6:4	FCLrnStage determines how to best most the coulomb-counter during the initial battery, since the capacity is not initially host can write the FullCAP and FullCA advance the FCLrnStage to 0x7 to sho behavior. Writing to 0x7, will enable the ing the normal MixRatio. If RdFCLrn is MiscCFG register, then FCLrnStage 7 FCLrn bits. At startup, FCLrnStage is reset to 0x0 a Fuel-gauge mixing time constant is 5.6 each 16% accumulation in Cycles, the doubled. The maximum allowed mixing set by NMIX in the SHFTCFG register. cumulation in Cycles, the FCLrnStage is The actual shift applied when mixing V to MixRemCap is the smaller of ((Cycle (NMIX+5)). For example, with NMIX = 13 (0xD) the constant is 12.8 hours and the progres the following: %Capacity Learned Mixing Time Constant 0% - 15% 5.625 seconds - Voltage FG dominal 16% - 31% 11.25 seconds - " 32% - 47% 22.5 seconds - " 48% - 63% seconds - " 64% - 79% minutes - " 80% - 95% minutes - " 96% - 111% (About 1 cycle) minutes - "	Il cyclii y know PNom rtcut the full massert will also and the 25 sectime continue c	ng of a vn. The values he star lixing used in to clear voltage onds. I constant constant constant ch 32% mente CAP +5 or mum ti oks lik ge >> 5 6 7 8 9 10 11	new es and tup is-he r the ge For t is is % ac-d. me e 45 1.5 3	
		_	rnCt-	ao		
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			2.0	٠		
				5		
FCI rnStage	6.4	_		6		
. OLinotage	ОТ		J	J		
			1	7		
			1	Ω	15	
			'	O	40	
		64% - 79%	2	9	1.5	
			0	40		
			2	10	3	
		milates	3	11	6	
		minutes - "				
		112% - 127%	3	12	12	
		minutes - " 128% - 143%	4	13	24	
		minutes - Coulomb-counter domi		.0		
		144% - 159%	4	14	48	
		minutes - " 160% - 175%	5	15	1.6	
		hours - " 176% - 191% hours - "	5	16	3.2	
		hours - " 192% - 207% (About 2 cycles)	6	17		
		6.4 hours - " 208% - 223% (And all future cycles 12.8 hours - "		18		

BITFIELD	BITS	DESCRIPTION	DECODE
SPR_3_3	3		
FillEmpty	2	FiltEmpty chooses whether empty is determined by a filtered or unfiltered voltage.	Set FiltEmpty = 1 to detect empty using AvgV-CELL. Set FiltEmpty = 0 to detect empty using VCELL, which can be only lightly filtered.
MixEn	1	MixEn enables mixing. Setting this to zero forces >>0.	
SPR	0		

FilterCfg (0x29)

BITFIELD	BITS	DESCRIPTION	DECODE
NEMPTY	15:14	Sets the filtering for empty learning for both the I_Avgempty and QRTable registers.	lavg_empty is learned with(NEMPTY+3) right shifts. QRTable is learned with (NEMPTY + sizeof(lavgempty) – sizeof(AvgCurrent)) right shifts.
NTEMP	13:11	Sets the time constant for the AverageTemperature register. The default POR value of 1h gives a time constant of 12min.	The equation setting the period is: AvergeTemperature time constant = 175.8ms x 2^(11 + TEMP)
NMIX	10:7	Sets the time constant for the mixing algorithm. The default POR value of Dh gives a time constant of 12.8 hours.	The equation setting the period is: Mixing Period = 175.8ms × 2(5+NMIX)
NAVGCELL	6:4	Sets the time constant for the AverageVCELL register. The default POR value of 2h gives a time constant of 45.0s.	The equation setting the period is: AverageVCELL time constant = 175.8ms × 2(6+NAVGVCELL)
NCURR	3:0	Sets the time constant for the AverageCurrent register. The default POR value of 4h gives a time constant of 11.25 seconds.	The equation setting the period is: AverageCurrent time constant = 175.8ms × 2^(2+NCURR)

RelaxCfg (0x2A)

The RelaxCFG register defines how the device detects if the cell is in a relaxed state. For a cell to be considered relaxed, current flow through the cell must be kept at a minimum while the change in the cell's voltage over time, dV/dt, shows little or no change. If AverageCurrent remains below the load threshold while VCELL changes less than the dV threshold over two consecutive periods of dt, the cell is considered relaxed.

BITFIELD	BITS	DESCRIPTION	DECODE
LoadThr	When current magnitude is less than Load- Thr, the device is considered to be unloaded. This is used to determine when to latch-load direction and VFSOC_EOL and when to start searching for relaxation. 8 bits. Stopload is detected when (RelaxCFG>>4)&0x0FE0) < [AvgCurrent8]. LoadThr does not include a sign bit.		The LSb is 5mA for $10m\Omega$ sense resistor.
dVThr	dVThr sets the relaxation criteria between VCELL and OCV. If the cell's voltage changes by less than dV over two consecutive periods set by dt, the cell is considered relaxed;		dVThr5 is 1.25mV per LSB. (1.25mV to 40mV range)
dTThr	3:0	dTThr configures the relaxation timer. Sets the time period over which change in VCELL is compared against dV. If the cell's voltage changes by less than dV over two consecutive periods set by dt, the cell is considered relaxed. The default value is 6 minutes.	The AvgVCELL is sampled at every dTThr interval (2dTThr x 0.1758s), and delta-V is checked by comparing to the previous sample.

MiscCfg (0x2B)

The MiscCFG control register enables various other functions of the device. The MiscCFG register default values should not be changed unless specifically required by the application.

- 0—Bit must be written 0. Do not write 1.
- 1—Bit must be written 1. Do not write 0.
- X—Don't Care. Bit may read 0 or 1.

BITFIELD	BITS	DESCRIPTION	DECODE
OopsFilter	15:12	OopsFilter default 3 (OopsFilter + 1) defines 0.39%/45s, 4%/6minutes, or 32%/hr rate limitation.	OopsFilter is 8%/hr per LSB.
EnBi1	11	Enable reset on battery-insertion detection. Set this bit to 1 to force a reset of the fuel gauge whenever a battery insertion is de- tected based on AIN pin monitoring. This bit is written to 1 at power-up.	
InitVFG	10	Setting this bit will cause the fuelgauge to re-initialize	
MixRate	9:5	Mixing Rate. This value sets the strength of the servo mixing rate after the final mixing state has been reached (> 2.08 complete cycles). The units are MR0 = $6.25\mu V$, giving a range up to $19.375mA$ with a standard $10m\Omega$ sense resistor. Setting this value to $00000b$ disables servo mixing and the IC continues with time-constant mixing indefinitely. The default setting is $18.75\mu V$ or $1.875mA$ with a standard sense resistor.	
RdFCLrn	4	Setting this bit will cause the FCLrn bits to be automattically cleared to 0 when the fule-gauge reaches FCLrnStage 7. This allows allows a more aggressive FullCapNom learning method to be used initially to quickly get an estimate of the battery capacity. Then the learning method will automatically revert to the most conservative method.	
vttl	3	This bit allows a lower voltage thermistor pullup.	
Vex	2	Set Vex=1 to disable voltage measurements. In this operation VCELL should be periodically written by the host.	
SACFG	1:0	SOC Alert CONFIG.	Set SACFG=0 to manage empty alerts using Rep-SOC. Set SACFG=1 to manage empty alerts using AvSOC. Set SACFG=10 to manage alerts using MixSOC. Set SACFG=11 to manage alerts using VFSOC.

TGain (0x2C)

The TGAIN and TOFF registers adjust the gain and offset of the temperature measurement A/D on the AIN pin to convert the result to a temperature value by the following equation: Temperature Register = (AIN Register ×TGAIN Register/16384) + (TOFF Register × 2).

Both these registers are signed two's complement. These registers allow for accurate temperature conversions when using a variety of external NTC thermistors.

BITFIELD	BITS	DESCRIPTION
TGAIN	15:0	TGAIN is a signed value with units of °C/64

TOff (0x2D)

The TGAIN and TOFF registers adjust the gain and offset of the temperature measurement A/D on the AIN pin to convert the result to a temperature value by the following equation: Temperature Register = (AIN Register ×TGAIN Register/16384) + (TOFF Register × 2).

Both these registers are signed two's complement. These registers allow for accurate temperature conversions when using a variety of external NTC thermistors.

BITFIELD	BITS	DESCRIPTION
TOFF	15:0	TOFF is a signed value with units of 2 ⁻⁷ in the LSb. Note that these units are 2X the TEMP register

CGain (0x2E)

The CGAIN and COFF registers adjust the gain and offset of the current measurement result. The current measurement A/D is factory trimmed to data-sheet accuracy without the need for the user to make further adjustments. The default power-up settings for CGAIN and COFF apply no adjustments to the Current register reading. For specific application requirements, the CGAIN and COFF registers can be used to adjust readings as follows:

Current Register = Current A/D Reading × (CGAIN Register/16384) + (2 × COFF Register)

For easiest software compatibility between systems, configure CGAIN to keep current LSb resolution at 0.15625mA. A minimum sense resistance of $5m\Omega$ is required due to the maximum range of CGAIN. This preserves resolution of current readings and capacities. Both these registers are signed two's complement. The default values of 4000h for CGAIN and 0000h for COFF preserve factory calibration and unit values ($1.5625\mu A/RSENSE$).

BITFIELD	BITS	DESCRIPTION
CGAIN	15:0	CGAIN is a signed value with units of 2-14. LSB is 0.0061%.

COff (0x2F)

BITFIELD	BITS	DESCRIPTION	DECODE
COFF	15:0	Current Offset	LSB is 3.15µA

QRTable20 (0x32)

BITFIELD	BITS	DESCRIPTION
QRTable20	15:0	QRTable is interpreted the same as BT07, using interpolation. However, during empty learning 2 QRTable registers should be modified with interpolation applied.

IAvgEmpty (0x36)

BITFIELD	BITS	DESCRIPTION	DECODE
lavg_empty	15:0	This is average of the current sampled at the last several empty events. lavgEmpty is used for defining a "pivot point" during empty learning, for adapting QRTable values. When empty occurs, if the current is much greater than lavgEmpty, then mostly gain is learned. But if the current is much lower than lavgEmpty, then mostly offset is learned.	lavgEmpty is a signed register with a 12.5μV/Rsns (1.25mA with 10mΩsense resistor)LSB in D3.

RComp0 (0x38)

BITFIELD	BITS	DESCRIPTION
SPR	15:8	
RCOMP0	7:0	This is the RCOMP value that is appropriate at Temperature = TempNom Adjust RCOMP0 as you would adjust RCOMP in ModelGauge 1.0 parts.

TempCo (0x39)

BITFIELD	BITS	DESCRIPTION
TempCoHot	15:8	Upper byte holds TempCoHot with units of 0.03125counts/°C.
TempCoCold	7:0	Lower byte holds TempCoCold with units of 0.125counts/°C

VEmpty (0x3A)

BITFIELD	BITS	DESCRIPTION	DECODE
V_Empty	15:7	Empty Voltage. Sets the voltage level for detecting empty.	A 10mV resolution gives a 0 to 5.11V range. This value is written to 3.12V at power-up.
V_Recover	6:0	Recovery Voltage. Sets the voltage level for clearing empty detection. Once the cell voltage rises above this point, empty voltage detection is reenabled.	A 40mV resolution gives a 0 to 5.08V range. This value is written to 3.68V at power-up.

QRTable30 (0x42)

BITFIELD	BITS	DESCRIPTION
QRTable30	15:0	QRTable is interpreted the same as BT07, using interpolation. However, during empty learning 2 QRTable registers should be modified with interpolation applied.

ConvgCfg (0x49)

The ConvgCFG register controls the converge to empty function of ModelGauge m5. This feature bends the RepSOC curve to closely match the voltage waveform as the SOC approaches empty, ensuring 0% is reported when V_Empty is reached.

BITFIELD	BITS	DESCRIPTION	DECODE
RepLow	15:12	Sets the threshold below which RepCap begins to bend upwards.	The LSB for RepLow is 2% and the range is 0% to 30%.
VoltLowOff	11:7	When the AvgVCELL drops below VoltLow, RepCap is bent downwards according to the ratio ((VCELL-Vempty)/VoltLowOff). The LSB for VoltLowOff is 20mV as the value here is a positive differential value to V_empty. Note that once this condition is reached, the RepLow math is replaced by this math.	
MinSlopeX	6:3	Sets the amount of slope-shallowing which occurs when RepSOC falls below RepLow.	MinSlopeX=1 corresponds to a 1/16 ratio, and MinSlopeX=15 corresponds to 15/16 ratio.
RepL_per_ stage	2:0	Adjusts the RepLow threshold by: RepL_per_stage*1%*(7-LearnStage). This allows the earlier learn-stages to effectively use a higher RepLow setting, while the final learn-stages are just set to RepLow.	The LSB for Rep_per_stage is 1% and the range is 0% to 7%.

Status2 (0xB0)

The Status 2 register indicates when the battery is full or empty.

BITFIELD	BITS	DESCRIPTION	DECODE
SPR_15_6	15:6		
FullDet	5	Fully_Charged.	Set FullDet = 1 VFSOC>99%; cleared if VF-SOC<98.5%. It means 'setting FullDet bit' is latched.
SPR_4_2	4:2		
Hib	1	The HCONFIG2 hib Bit is copied to STATUS2 for customer application purposes. Hib indicates whether or not the FG is in hibernate mode.	
SPR_0	0		

TTF_CFG (0xB5)

BITFIELD	BITS	DESCRIPTION
SPR	15:3	
TTF_CFG	2:0	Configures the filtering rate for learning CV halftime for TTF calculation.

CV_MixCap (0xB6)

BITFIELD	BITS	DESCRIPTION
CV_MixCap	15:0	The MixCapacity when CV mode has been observed to begin.

CV_HalfTime (0xB7)

BITFIELD	BITS	DESCRIPTION
CV_Halftime	15:0	CV_HalfTime is the observed half-time from CV_Start until AvgCurrent <cc_ Current/2. It is the exponential decay half-life of the current curing CV mode charging.</cc_

CGTempCo (0xB8)

BITFIELD	BITS	DESCRIPTION
CGTempCo	15:0	CGTempCo configures the temperature coefficient for metal resistance, if the calculation is enabled. CGTempCo is restored from nCGTempCo OTP if enabled. If metal resistance is enabled, but OTP restore for CGTempCo is not enabled, then CGTempCo gets loaded with a good default value for using copper (0x20C8 for 8%/20degC).

Curve (0xB9)

The CURVE register is used to enable thermistor curvature adjustment. This allows the temperature to be calculated accurately over a wider range than just using TGAIN and TOFF.

BITFIELD	BITS	DESCRIPTION	DECODE
ECURVE	15:8	Ground Resistance Thermistor Compensation. Compensates the thermistor calculation by considering the battery current through the BATT- path and BATT- connector. Normally set to 0 unless the application has high currents or significant GND resistance between chip GND and thermistor GND.	LSB is $0.9765625m\Omega$, so the range is 0 to $249.0234375m\Omega$.
TCURVE	7:0	Sets the curvature compensation for thermistor calculations if thermistor curvature compensation is enabled. OTPCFG1.enCRV = 1. When enabled, the temperature translation is modified according to the following equation: (original temp translation equation): TEMPx = (AIN x TGAIN)/0x4000 + TOFF x 2 Combined with new translation work: TEMP = TEMPx + TCURVE x (TEMP-0x1400) x (TEMP-0x1400) /0x40000 x if(Temp<0x1400,-1,1).	

Config2 (0xBB)

BITFIELD	BITS	DESCRIPTION	DECODE
SPR	15:11		
FCThmHot	10	Set to 1 to enable thermistor hot forcedly, regardless of actually sense thermistor temperature.	
ThmHotEn	9	Set to 1 to enable ThmHot function by comparing Voltage/Temp condition with THMHOT (40h). ThmHot function is FG charger input current-limit control.	
ThmHo- tAlrtEn	8	Set to 1 to enable ThmHotAlrtEn alert. If Thm-HotAlrtEn = 0, the alert-on is disabled.	
dSOCen	7	Set to 1 to enable SOC 1% change alert. If dSCOCen = 0, the alert-on is disabled.	
TAIrtEn	6	Set to 1 to enable temperature alert. If TAIrtEn = 0, the alert-on is disabled.	
LdMdl	5	Host sets this bit to 1 in order to initiate firmware to finish processing a newly loaded model. Firmware clears this bit to zero to indicate that model loading is finished.	
OCVQen	4	Set OCVQen=1 to enable automatic empty compensation based on VFOCV information.	
ISysNCurr	3:0	Sets the time constant for the AvgISys register. The default POR value of 0100b gives a time constant of 5.625. The equation setting the period is: AvgISys time constant = 45s x 2(ISysNCurr-7).	Sets the time constant for the AvgISys register. The default POR value of 0100b gives a time constant of 5.625. The equation setting the period is: AvgISys time constant = 45s x 2(ISysNCurr-7).

RippleCfg (0xBD)

BITFIELD	BITS	DESCRIPTION	DECODE
kDV	15:3	Sets the corresponding amount of capacity to compensate proportional to the ripple. kDV contributes empty compensation as a function of ripple voltage.	The LSB units for kDV are 0.05/256 %/mV.
NR	2:0	Sets the filter magnitude for ripple observation.	Set NR[2:0] for the below filter timing (i.e., Ripple Time Range): 0x0 = 1.4s 0x1 = 2.8s 0x2 = 5.6s 0x3 = 11.2s 0x4 = 22.4s 0x5 = 45s 0x6 = 90s 0x7 = 3min

Measurement Registers

Register Details

Temp (0x08)

BITFIELD	BITS	DESCRIPTION	DECODE
TEMP	15:0	This is the most recent trimmed temperature measurement. Temperature is measured every 1.4 seconds.	When using AIN for temperature (Tex = 0), configure TGAIN and TOFF to adjust the AIN measurement to provide units degrees in the high-byte of Temp. When TGAIN and TOFF are configured properly for the selected thermistor, the LSB is 0.0039°C and the upper byte has units 1°C. Temp is a signed register. To configure the BT07 to receive temperature information from the I ² C, set Tex = 1 and periodically write the Temp register with the appropriate temperature.

Vcell (0x09)

BITFIELD	BITS	DESCRIPTION	DECODE
VCELL	15:0	This is the most recent trimmed cell voltage result. It represents an FIR average of raw results. The VOLT_Raw is sampled every 175.8ms and gain and offset trim are applied to calculate VCELL.	Bits D15 to D0 represent that 15-bit conversion result. VCELL has 78.125μV per LSB.

Current (0x0A)

BITFIELD	BITS	DESCRIPTION
Current	15:0	The IC measures the voltage between the CSP and CSN pins and the result is stored as a two's complement value in the Current register. Voltages outside the minimum and maximum register values are reported as the minimum or maximum value. The register value should be divided by the sense resistance to convert to Amperes. The value of the sense resistor determines the resolution and the full-scale range of the current readings.

AvgCurrent (0x0B)

BITFIELD	BITS	DESCRIPTION	DECODE
AvgCurrent	15:0	This is the 0.7s to 6.4hr (configurable) IIR average of the current. This register represents the upper 16 bits of the 32-bit shift register that filters current. The average should be set equal to Current upon startup.	Units of LSbit are 1.5625µA/Rsns.

AvgTA (0x16)

BITFIELD	BITS	DESCRIPTION	DECODE
AvgTA	15:0	This is the 6min to 12hr (configurable) IIR average of the Temperature. The average is set equal to Temp upon startup.	Units of LSbit are 0.0039°C. The upper byte has units 1°C.

AvgVCeII (0x19)

BITFIELD	BITS	DESCRIPTION	DECODE
AvgVCELL	15:0	This reports the 12s to 24min (configurable) IIR average of VCELL.The average is set equal to VCELL at startup.	AvgVCELL has 78.125uV per LSB

MaxMinTemp (0x1A)

BITFIELD	BITS	DESCRIPTION	DECODE
MaxTempera- ture	15:8	Records the maximum Temperature.	Units of LSB are 1°C.
MinTempera- ture	7:0	Records the minimum Temperature.	Units of LSB are 1°C.

MaxMinVolt (0x1B)

BITFIELD	BITS	DESCRIPTION	DECODE
MaxVoltage	15:8	Records the VCELL maximum voltage.	Units of LSB are 20mV.
MinVoltage	7:0	Records the VCELL minimum voltage.	Units of LSB are 20mV.

MaxMinCurr (0x1C)

BITFIELD	BITS	DESCRIPTION	DECODE
MaxCharge- Current	15:8	Records the maximum charge current.	8-bit values with 0.4mA resolution.
MaxDisCur- rent	7:0	Records the maximum discharge current.	8-bit values with 0.4mA resolution.

AIN0 (0x27)

BITFIELD	BITS	DESCRIPTION	DECODE
AIN0	15:0	This is the most recent trimmed ratiometric AIN0 measurement, which is generally used for measuring temperature. AIN0 is measured every 1.4 seconds (if Ten=1).	AIN0 is an unsigned register where 0xFFFF indicates 100% ratio between AIN0/THRM. LSb is 2-16

AtTTF (0x33)

BITFIELD	BITS	DESCRIPTION
AtTTF	15:0	

Timer (0x3E)

BITFIELD	BITS	DESCRIPTION
TIMER	15:0	Timer increments once every task period. With default TaskPeriod, timer has units 0.1758 seconds. The timer manages the following tasks: 1) Thermistor measurements occur once every 8 tasks. 2) Debouncing repeats for 8 TIMER ticks. 3) dV is measured based on dTthr TIMER ticks.

ShdnTimer (0x3F)

The SHDNTIMER register sets the timeout period from when a shutdown event is detected until the device disables the LDO and enters low-power mode.

BITFIELD	BITS	DESCRIPTION
SHDN_THR	15:13	Sets the shutdown timeout period from a minimum of 45s to a maximum of 1.6h. The default POR value of 7h gives a shutdown delay of 1.6h. The equation setting the period is: Shutdown Timeout Period = 175.8ms × 2(8+THR)
SHDNCTR	12:0	Shutdown Counter. This register counts the total amount of elapsed time since the shutdown trigger event. This counter value stops and resets to 0 when the shutdown timeout completes. The counter LSb is 1.4s.

QH0 (0x4C)

BITFIELD	BITS	DESCRIPTION
QH0	15:0	Last sampled QH for dQ accumulation.

VRipple (0xBC)

BITFIELD	BITS	DESCRIPTION	DECODE
Vripple	15:0	It is for the voltage compensation on battery capacity report. It is for the internal usage only.	LSB unit = 1.25/16mV.

TimerH (0xBE)

BITFIELD	BITS	DESCRIPTION
TIMERH	15:0	TIMERH is a 16-bit high-word extension to the TIMER register. This extension allows time counting up to 24 years. This register can be enabled in the save and restore registers.

ModelGauge m5 Output Registers

Register Details

RepCap (0x05)

BITFIELD	BITS	DESCRIPTION	DECODE
RepCap	15:0	AvCap is modified to prevent from reporting any sudden changes in capacity when there is a sudden load change. RepCap forecasts the same time-to-empty as AvCap.	$0.5 mAh$ per LSB with $10 m\Omega$ sense resistor.

RepSOC (0x06)

BITFIELD	BITS	DESCRIPTION	DECODE
RepSOC	15:0	RepSOC is the complete calculation for state-of-charge. This includes all processing, including: ModelGauge mixing, and empty compensation.	16 bit result. The high byte indicates 1% per LSB. The low byte reports fractional percent.

Age (0x07)

BITFIELD	BITS	DESCRIPTION	DECODE
Age	15:0	Age represents the percentage of full capacity relative to the design capacity.	The high byte is 1%/LSB, and the low byte is 1/256 %/LSB. Age is an unsigned int. Equation is FullCap / DesignCap x 100%

QResidual (0x0C)

BITFIELD	BITS	DESCRIPTION	DECODE
Qresidual	15:0	This is the capacity which is not available because of the battery impedance and load current. This value changes when the load current or temperature changes. This is based on a learned parameter.	16-bit value. 0.5mAh per LSB with $10m\Omega$ sense resistor.

MixSOC (0x0D)

BITFIELD	BITS	DESCRIPTION	DECODE
MixSOC	15:0	SOC is the remaining state-of-charge in the battery, including capacity that may be unavailable because of the discharge rate.	16 bit result. The high byte indicates 1% per LSB. The low byte reports fractional percent.

AvSOC (0x0E)

BITFIELD	BITS	DESCRIPTION	DECODE
AvSOC	15:0	AvSOC is the available state-of-charge. This includes all processing, including: Model-Gauge mixing, and empty compensation.	16 bit result. The high byte indicates 1% per LSB. The low byte reports fractional percent.

MixCap (0x0F)

BITFIELD	BITS	DESCRIPTION	DECODE
МіхСарН	15:0	This is the MS word representing remaining capacity with coulomb-counter + voltage-fuel-gauge mixing. This does not include empty compensation, and includes any capacity which may be unavailable because of the discharge rate. Current results are accumulated every 175.8ms.	0.5mAh per LSB with 10 m $Ω$ sense resistor.

FullCap (0x10)

BITFIELD	BITS	DESCRIPTION	DECODE
FullCAP	15:0	This register holds the temperature compensated full capacity value. This also compensates for the temperature dependence of charge termination. The FullCapNom valus is multiplied by a temperature correction factor (FCTC) and the result is stored in this register.	$0.5 mAh$ per LSB with $10 m\Omega$ sense resistor.

TTE (0x11)

BITFIELD	BITS	DESCRIPTION	DECODE
hr	15:10	Remaining time-to-empty is calculated as (AvCap) / AvgCurrent. When enabled, the AtRate value is substituted for AvgCurrent in this calculation.	LSB unit = 1.6hr
mn	9:4	Remaining time-to-empty is calculated as (AvCap) / AvgCurrent. When enabled, the AtRate value is substituted for AvgCurrent in this calculation.	LSB unit = 1.5mn
sec	3:0	Remaining time-to-empty is calculated as (AvCap) / AvgCurrent. When enabled, the AtRate value is substituted for AvgCurrent in this calculation.	LSB unit = 5.625s

Rslow (0x14)

BITFIELD	BITS	DESCRIPTION	DECODE
RSLOW	15:0	This reports the battery's slow internal resis-	16 bit value. Units of LSbit are 2-12Ω with 10mΩ
		tance.	sense resistor.

Cycles (0x17)

BITFIELD	BITS	DESCRIPTION	DECODE
Cycles	15:0	Odometer style accumulation of battery cycles.	The LSB indicates 1% of a battery cycle (1% charge + 1% discharge). One cycle (Cycles = 100%) indicates 100% charge and discharge.

AvCap (0x1F)

BITFIELD	BITS DESCRIPTION		DECODE
AvCap	15:0	This is the remaining capacity with coulomb- counter + voltage-fuel-gauge mixing, after accounting for capacity that is unavailable due to the discharge rate.	16-bit value. 0.5mAh per LSB with $10m\Omega$ sense resistor.

TTF (0x20)

BITFIELD	BITS	DESCRIPTION	DECODE
hr	15:10	Remaining time-to-full as calculated by firm-ware.	LSB unit = 1.6hr
mn	9:4	Remaining time-to-full as calculated by firm-ware.	LSB unit = 1.5mn
sec	3:0	Remaining time-to-full as calculated by firm-ware.	LSB unit = 5.625s

FullCapRep (0x35)

BITFIELD	BITS	DESCRIPTION	DECODE
FullCapRep	15:0	FullCapRep reports the full capacity using the MAX17047 method, where full gets defined as "the RepCap when charge termination occurs". This is different from the new method (see FullCap register) in which FullCap is automatically compensated when the temperature and load condition changes. I.e., if the temperature is reduced, FullCap suddenly reduces (not waiting for any charging or charge termination), while FullCapRep does not update until the next charge termination.	$0.5 \text{mAh per LSB with } 10 \text{m}\Omega$ sense resistor.

FCTC (0x37)

BITFIELD	BITS	DESCRIPTION	DECODE
FCTC	15:0	This temperature correction factor is used to calculate the FullCap0 as a function of FullCapNom and the present temperature (TA).	Units are 1/2 ²¹ C.

dQAcc (0x45)

BITFIELD	BITS	DESCRIPTION	DECODE
dQacc	15:0	For continuous learning and zigzag learning, this is an always increasing number. I.e., as the battery is charged, this register increases. As it is discharged, it also increases.	Units of LSb are 16mAh with $10m\Omega$ sense resistor. Max range is 8X RemCap register.

dPAcc (0x46)

BITFIELD	BITS	DESCRIPTION	DECODE	
dPacc	15:0	Similar behavior as dQacc, except this is derived from the VF_SOC.	16 bit value. Units of LSb are 0.015625% (1/64 %).	

VFRemCap (0x4A)

BITFIELD	BITS	DESCRIPTION
VFRemCap	15:0	Remaining Capacity according to the voltage fuel-gauge.

MaxError (0xBF)

BITFIELD	BITS	DESCRIPTION	DECODE
MaxError	15:0	At each empty learning event, it is a filter for the new error. It is for internal usage only.	LSB unit = 1/256%.

AtQresidual (0xDC)

BITFIELD	BITS	DESCRIPTION	
AtQresidual	15:0	AtQresidual is calculated as normal Qresidual, except using AtRate instead of current.	

AtTTE (0xDD)

BITFIELD	BITS	DESCRIPTION	DECODE
AtTTE	15:0	Calculated time-to-empty based on the formula AtTTE = ((AtAvCap,0) / AtRate) >> 5	LSB unit = 5.625s.

AtAvSOC (0xDE)

BITFIELD	BITS	DESCRIPTION	
AtAvSOC	15:0	AtAvSOC is calculated as normal AvSOC, except using AtRate instead of current.	

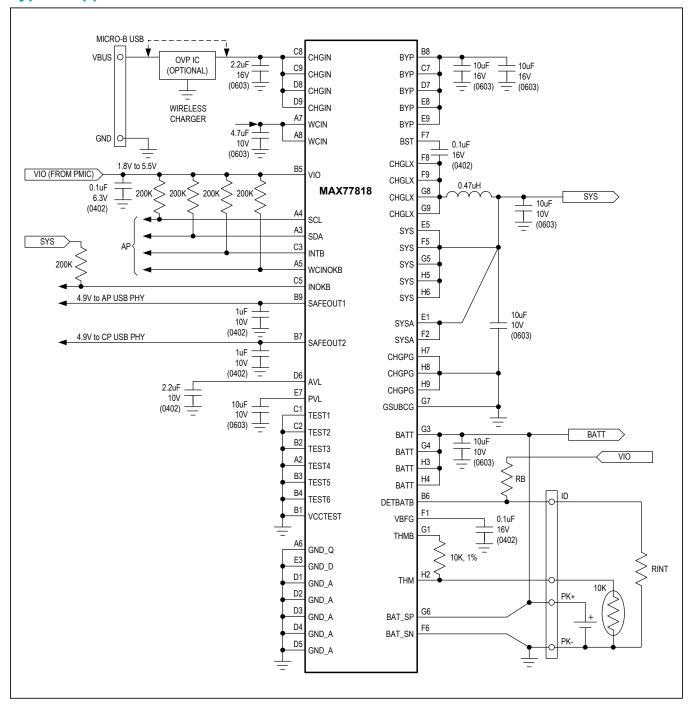
AtAvCap (0xDF)

BITFIELD	BITS	DESCRIPTION	
AtAvCap	15:0	AtAvCap is calculated as normal AvCap, except using AtQresidual instead of Qresidual.	

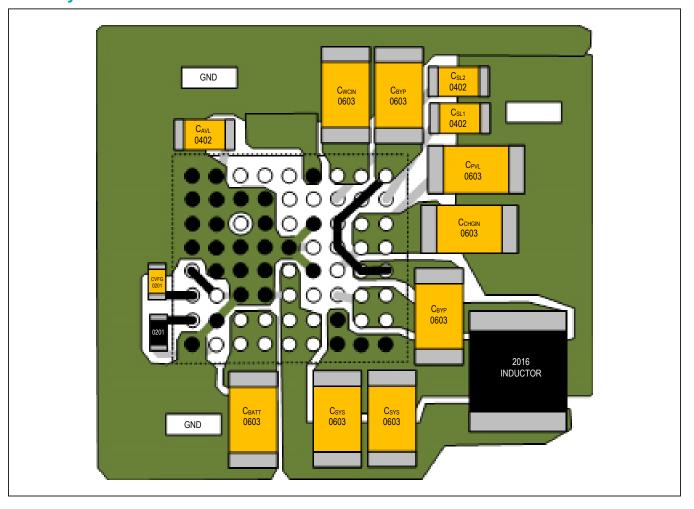
VFOCV (0xFB)

BITFIELD	BITS	DESCRIPTION
VFOCV	15:0	

Typical Application Circuit



PCB Layout Guide



Layout Guidelines

Good PCB layout is essential to minimize ground bounce and EMI and achieve voltage, temperature, and current measurement accuracy. Follow these guidelines:

- 1) Connect the power ground plane to the analog ground plane with a star connection.
- Place CHGIN, WCIN, BYP, SYS, and BATT bypass capacitors as close as possible to the IC pins and connect them to the power ground plane on the PCB top layer.
- Use wide and short traces for high current connections such as CHGIN, WCIN, SYS, and BATT.
- 4) All bypass capacitors should be placed as close to the chip as possible, connect bypass capacitors to the closest ground plane.
- 5) BAT_SP and BAT_SN should have direct kelvin sensing connections to the interface of battery.
- The kelvin traces should not be shared with other circuits and vias on kelvin traces are not recommended.

Ordering Information

PART	TEMP RANGE	PIN-PACKAGE
MAX77818EWZ+	-40°C to +85°C	72-pin WLP 0.4mm pitch, 3.867mm x 3.608mm (±0.015mm X and Y)

⁺Denotes lead(Pb)-free/RoHS-compliant package.

Chip Information

PROCESS: BICMOS

Package Information

For the latest package outline information and land patterns (footprints), go to www.maximintegrated.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE	PACKAGE	OUTLINE	LAND
TYPE	CODE	NO.	PATTERN NO.
72 WLP	W723B3+1	21-0477	Refer to Application Note 1891

MAX77818

Dual Input, Power Path, 3A Switch Mode Charger with Fuel Gauge

Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	6/15	Initial release	_
1	10/15	Corrected Figure 5 and minor errors	19, 22, 27, 36
2	2/16	Corrected m5 designation throughout, updated Note 7, <i>Safeout LDO</i> section, <u>Figure 2</u> , and <u>Figure 5</u>	1, 16, 21, 23
3	9/17	Corrected errors, added fuel gauge descriptions	1–102
4	3/20	Updated Typical Operating Circuit, Safeout LDOs Electrical Characteristics table, and Figure 1	1, 18, 19, 28

For pricing, delivery, and ordering information, please visit Maxim Integrated's online storefront at https://www.maximintegrated.com/en/storefront/storefront.html.

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PI3VST01ZEEX PI5USB1458AZAEX PI5USB1468AZAEX MCP16502TAC-E/S8B MCP16502TAE-E/S8B MCP16502TAA-E/S8B

MCP16502TAB-E/S8B TCKE712BNL,RF ISL91211AIKZT7AR5874 ISL91211BIKZT7AR5878 MCP16501TC-E/RMB ISL91212AIIZ
TR5770 ISL91212BIIZ-TR5775 CPX200D AX-3005D-3 TP-1303 TP-1305 TP-1603 TP-2305 TP-30102 TP-4503N MIC5167YML-TR

LPTM21-1AFTG237C LR745N8-G MPS-3003L-3 MPS-3005D SPD-3606 STLUX383A TP-60052 ADN8834ACBZ-R7 LM26480SQ
AA/NOPB LM81BIMTX-3/NOPB LM81CIMT-3/NOPB MIC5166YML-TR GPE-4323 GPS-2303